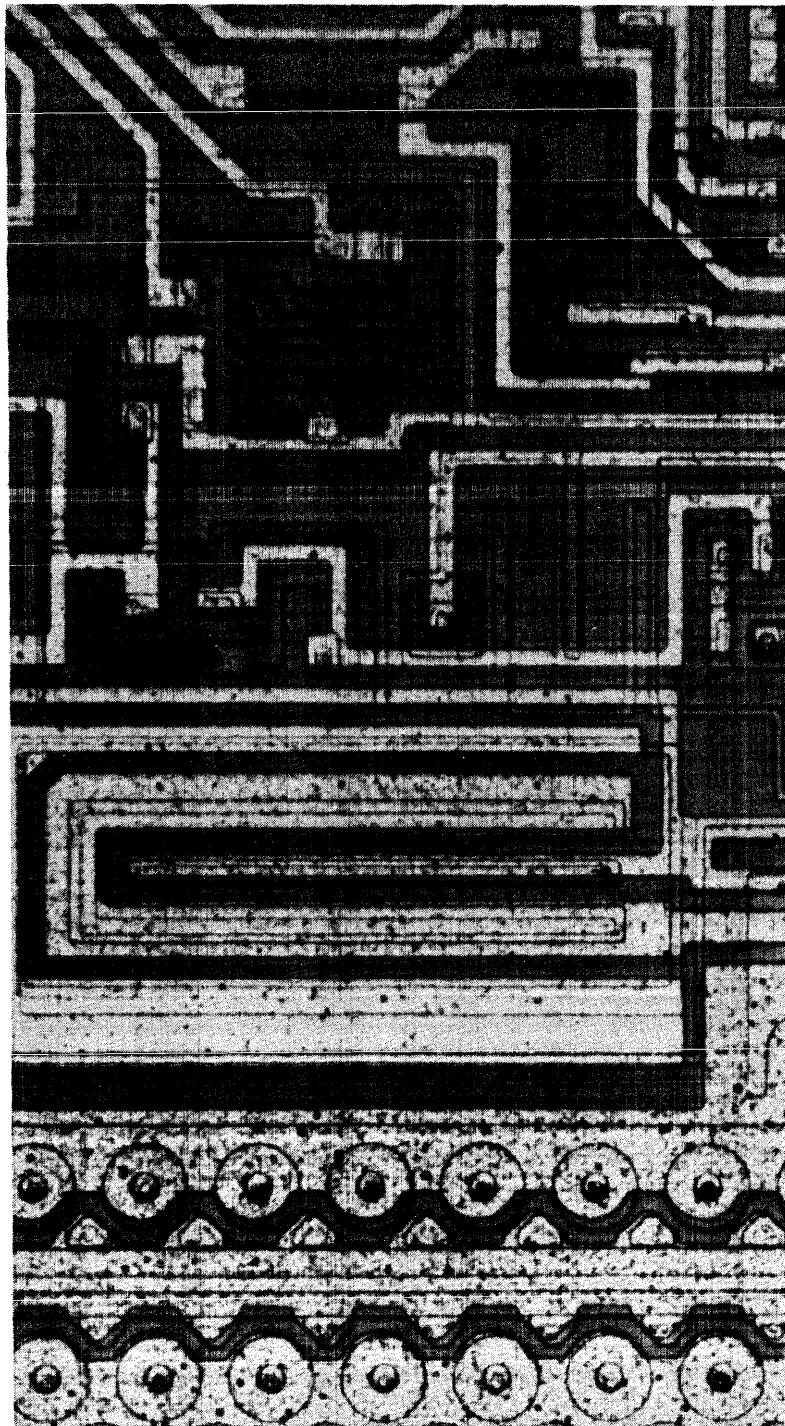


1984

# IC MASTER

VOLUME II



Courtesy of Intel Corporation

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INTERFACE—Analog to Digital Converters

Bits Res.	Linearity Error $\pm$ LSB	Conversion Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Integrating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																		
3	1/8	0.02	—	x												ADC-HU3B	Datel (2862)	
4	1/4	0.033 *	250	x		x		x			x				x	TDC1021	† TRW-LSI (3704)	
				x		x		x		x		x		TDC1021	TRW-LSI			
	1/2	0.03 *	800	x		x									x	UAB1005	Thomson-CSF	
				x		x							x		UAC1005	† Thomson-CSF		
6	—	0.004 *	600 *	x											x	54F500	† Fairchild (762)	5
				x										x		74F500	Fairchild (762)	
	1	—	—										x		MAH-0801	AD (2839)		
	1/8	—	—													AD5010K	AD (2839)	
																AD6020K	AD (2839)	10
	1/4	0.02	450	x	x						x				x	SDA5010	Siemens	
				x	x							x		x		SDA6020	Siemens	
	1/2	0.01	1600	x	x			x							x	TDC1029	† TRW-LSI	
				x														ZN440CJ
		0.033 *	1000	x		x		x			x				x	TDC1014	† TRW-LSI (3704)	15
				x		x		x		x		x		x		TDC1014	TRW-LSI	
		0.066	315										x	x	CA3300	RCA (3600)		
	90	120		x											x	NE5036	Signetics	
	100	100		x											x	NE5037	Signetics (3610)	
	100MHz	—		x	x			x							x	TDC1029	† TRW-LSI (3704,3707)	20
	1000	20 *		x											x	TL507C	TI	
7	1/4	—	1200	x											x	MC10315L	Motorola (3070)	
				x											x	MC10317L	Motorola (3070)	
	1/2	0.055	1604	x	x			x							x	TDC1027	† TRW-LSI	
				x	x			x							x	TDC1027	† TRW-LSI (3704,3706)	25
	1	—	100 *											x	TM1070	Telmos		
	6 MHz	150		x										x	TML1073	Telmos		
8	1/4	9	—		x	x							x	x		ZN447E-8	DDC	30
					x	x							x	x		ZN447J-8	† DDC	
		125		x	x							x	x		ADC-847B	Datel (2863)		
				x	x							x	x		ADC-847M	† Datel (2863)		
		125 *		x	x							x	x		ZN447E	Ferranti		
				x	x							x	x		ZN447J	† Ferranti		
	100 *	875		x										x	ADC0801	† Intersil (2993)	35	
				x										x	ADC0801C	Intersil (2993)		
	110	875		x										x	ADC0801	† National		
				x										x	ADC0801C	National		
	200	0.3 *		x						x					TL522	TI		
	1/2	0.02	4500	x	x			x							x	TDC1025	† TRW-LSI	40
				x				x		x		x		x		TDC1007	† TRW-LSI (3704)	
		0.033 *	2500	x		x		x			x			x	TDC1007	TRW-LSI		
		0.05	1320	x											x	TDC1048	† TRW-LSI (3703)	
		0.066	315										x	x	CA3306	RCA (3600)		
	0.4	400		x											x	TDC1001	TRW-LSI (3704)	45
	0.6	1250		x	x			x							x	ADC-815MC	Datel	
				x	x			x		x		x	x		ADC-815MM	† Datel		
						x	x	x					x	x	MN5815	† MicroNet (3043)		
	0.9	1400	1550			x	x						x	x	x	DDC-5101	† DDC	50
				x	x									x	x		ADC-5101	Datel (2860)
				x	x									x	ADC-5101E	Datel (2860)		
				x		x									x	ADC-5101H	† Datel (2860)	
				x		x	x						x	x	x	MN5101	MicroNet (3043)	
				x		x	x						x	x	x	MN5101H	† MicroNet (3043)	

Bin.—Binary      Compl.—Complementary      CTC—Compl. 2's Compl.      Mux. In.—Multiplexed Inputs      Par. Out.—Parallel Output  
 Off.—Offset      Magn.—Magnitude      Int. Ref.—Internal Reference      S&H—Sample and Hold      Ser. Out.—Serial Output



## ANALOG TO DIGITAL CONVERTERS

MODEL NO.	RESOLUTION	CONVERSION TIME	LINEARITY	OPERATING TEMP. RANGE (°C)
ADC-826MR	10 bits	1.4 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-827MC	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-827MR	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-830C	8 bits	100 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-833R	6 bits	66 nsec	$\pm 1/2$ LSB	-25 to +85
<b>NEW</b> ADC-847A	8 bits	9 $\mu$ sec	$\pm 1$ LSB	0 to +70
<b>NEW</b> ADC-847B	8 bits	9 $\mu$ sec	$\pm 1/4$ LSB	0 to +70
<b>NEW</b> ADC-847M	8 bits	9 $\mu$ sec	$\pm 1/4$ LSB	-55 to +125
ADC-856C	10 bits	1 $\mu$ sec/LSB	$\pm 1/2$ LSB	0 to +70
ADC-856M	10 bits	1 $\mu$ sec/LSB	$\pm 1/2$ LSB	-55 to +125
ADC-5101	8 bits	900 nsec	$\pm 1/2$ LSB	0 to +70
ADC-5101E	8 bits	900 nsec	$\pm 1/2$ LSB	-25 to +85
ADC-5210	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5210E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-5211	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5211E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-5212	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5212E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-5213	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5213E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-5214	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5214E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-5215	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5215E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
ADC-5216	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
ADC-5216E	12 bits	13 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85

## DIGITAL TO ANALOG CONVERTERS

MODEL NO.	RESOLUTION	OUTPUT SETTLING TIME	LINEARITY	OPERATING TEMP. RANGE (°C)
DAC-HA10BC	10 bits	1.3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HA10BR	10 bits	1.3 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-HA12BC	12 bits	5 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HA12BR	12 bits	5 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-HA14BC	14 bits	7 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HA14BR	14 bits	7 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-HF8BMC	8 bits	25 nsec	$\pm 1/2$ LSB	0 to +70
DAC-HF8BMR	8 bits	25 nsec	$\pm 1/2$ LSB	-25 to +85
DAC-HF10BMC	10 bits	25 nsec	$\pm 1/2$ LSB	0 to +70
DAC-HF10BMR	10 bits	25 nsec	$\pm 1/2$ LSB	-25 to +85
DAC-HF12BMC	12 bits	50 nsec	$\pm 1/2$ LSB	0 to +70
DAC-HF12BMR	12 bits	50 nsec	$\pm 1/2$ LSB	-25 to +85
DAC-HK12BGC	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HK12BMC	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HK12BMR	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-HP16BGC	16 bits	15 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HP16BMC	16 bits	15 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HP16BMR	16 bits	15 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-HZ12BGC	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HZ12BMC	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-HZ12BMR	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-IC8BC	8 bits	300 nsec	$\pm 1/2$ LSB	0 to +70
DAC-IC8BM	8 bits	300 nsec	$\pm 1/2$ LSB	-55 to +125
DAC-IC10BC	10 bits	250 nsec	$\pm 1$ LSB	0 to +70
DAC-IC10B	10 bits	250 nsec	$\pm 1/2$ LSB	0 to +70
DAC-IC10BM	10 bits	250 nsec	$\pm 1/2$ LSB	-55 to +125
DAC-UP8BC	8 bits	2 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-UP8BM	8 bits	2 $\mu$ sec	$\pm 1/2$ LSB	-55 to +125
DAC-UP10BC	10 bits	5 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-08BC	8 bits	150 nsec	$\pm 1/2$ LSB	0 to +70
DAC-08BM	8 bits	150 nsec	$\pm 1/2$ LSB	-55 to +125
DAC-85C-CBI-I	12 bits	300 nsec	$\pm 1/2$ LSB	0 to +70
DAC-85C-CBI-V	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-85-CBI-I	12 bits	300 nsec	$\pm 1/2$ LSB	-25 to +85
DAC-85-CBI-V	12 bits	3 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
DAC-562C	12 bits	400 nsec	$\pm 1/2$ LSB	0 to +70
DAC-562M	12 bits	400 nsec	$\pm 1/2$ LSB	-55 to +125
DAC-608	8 bits	1 $\mu$ sec	$\pm 1/2$ LSB	0 to +70
DAC-610	10 bits	500 nsec	$\pm 1/2$ LSB	0 to +70
DAC-612	12 bits	1 $\mu$ sec	$\pm 1/2$ LSB	-25 to +85
<b>NEW</b> DAC-7523	8 bits	200 nsec	$\pm .1\%$	0 to +70
<b>NEW</b> DAC-7533	10 bits	800 nsec	$\pm .01\%$	0 to +70
<b>NEW</b> DAC-7541	12 bits	1 $\mu$ sec	$\pm .012\%$	0 to +70

DATEL 11 CABOT BOULEVARD, MANSFIELD, MA 02048/TEL. (617) 339-9341/TWX 710-346-1953/TLX 951340



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IC Updates**

**These quarterly updates  
keep your IC Master  
current the year  
'round.**

**Look for them  
in Integrated  
Circuits Magazine**



**MITEL**

# MD54/74HCT245R

Octal Bus Transceivers with 3-State Buffered Outputs  
Preliminary Information

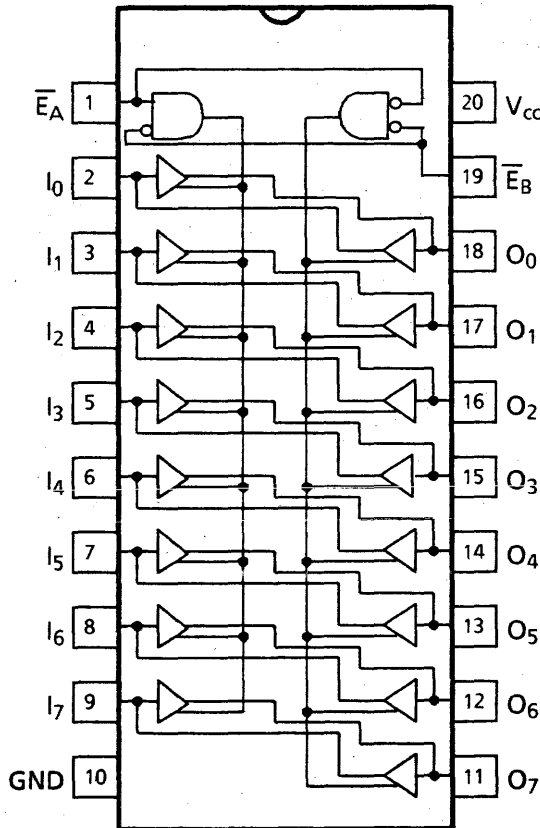
October '83

## Features

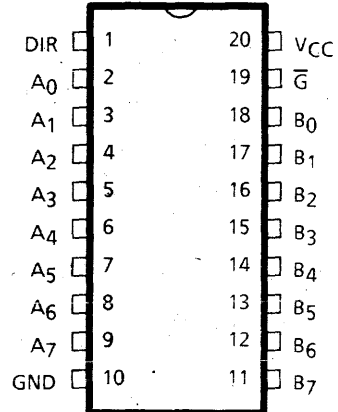
- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS245 types

## Description

These ISO-CMOS Octal line driver/buffers are designed for high speed asynchronous two-way communication between data buses. The control function inputs minimize external timing requirements. Data transmission from the A bus to the B bus, or from the B bus to the A bus, is provided depending upon the logic level; at the direction control input (DIR) pin. The enable input (G) pin can be used to disable the device outputs so that the buses are effectively isolated from each other.



CONNECTION DIAGRAM  
DIP (TOP VIEW)



## Ordering Information

MD54HCT245RCB, Cerdip  
-55°C to 125°C  
with MIL 883B option

MD74HCT245RE, Plastic Dip  
-40°C to 85°C

PIN	DESCRIPTION
A <sub>0</sub> - A <sub>7</sub>	Bus A, Data Inputs/Outputs
B <sub>0</sub> - B <sub>7</sub>	Bus B, Data Inputs/Outputs
DIR	Direction Control Pin
$\bar{G}$	Enable Input, Active Low
V <sub>CC</sub>	Supply Voltage
GND	System Ground

FUNCTION TABLE

INPUT	OUTPUT	
ENABLE (G)	CONTROL (DIR)	OPERATION
L	L	A to B
L	H	A to B
H	X	ISOLATION

H = logic "1", L = logic "0", X = don't care,  
Z = high impedance

1984

# IC MASTER

VOLUME II

Engineering design begins with the IC MASTER

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Technical Director, IC Division, **Jerry Metzger**

Assistant Editors, **Addie Bisignano,**  
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Manager, Catalog Production, **Lanny Levin**

Catalog Design Director, **Michelle Arnold**

Assistant Catalog Design Director, **Bill Hennessey**

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**HEARST BUSINESS COMMUNICATIONS, INC./**  
**UTP DIVISION**

**East Coast:**

645 Stewart Avenue

Garden City, NY 11530

Tel: (516) 222-2500 Telex II 510-222-1673 (UTP GRCY)

**West Coast:**

1333 Lawrence Expressway, Suite 101

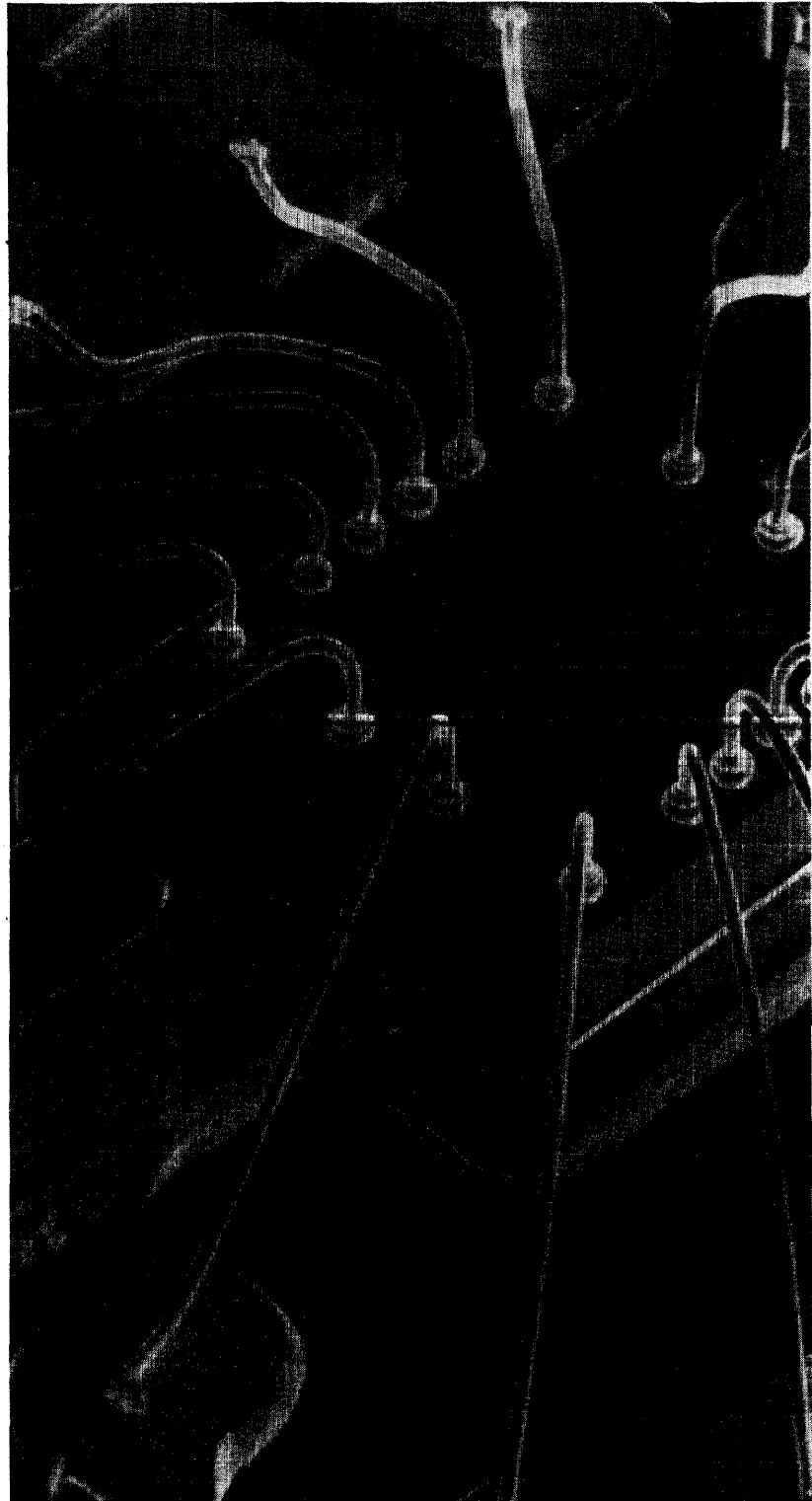
Santa Clara, CA 95051

Tel: (408) 248-8044 Telex II 910-338-0272

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Courtesy of Delco Electronics

# IMPORTANT FEATURES OF YOUR IC MASTER

## ONE COMPLETE SOURCE

IC MASTER is the original and only complete guide to currently available integrated circuits, microcomputer boards, development systems, PROM programmers, gate arrays, and other related components of concern to the design engineer. It has become the first place to look in the critical selection of ICs, boards, systems, and equipment. If only one device can fit the requirements of a new design or if hundreds are available, you can find out in seconds by using the IC MASTER.

## EASY TO USE

The IC MASTER saves you time. No longer do you have to spend long, tedious hours and days searching through manufacturers' catalogs and data sheets for information. The MASTER gives you—at your fingertips—an easy way to narrow your IC choices quickly, accurately, and systematically with the knowledge that you have just surveyed the entire industry.

## PART NUMBER INDEX

This revolutionary index lists all device types made by over 225 manufacturers in numerical sequence excluding prefixes or suffixes. You can find a device number even though you do not know either the full part number or even the manufacturer. Once a basic device number is located in the index, you can obtain instant identification of all manufacturers making a device by that number, regardless of function, and determine the full part number designation. All page references to data sheet material and any existing application note abstracts are also provided. The Part Number Index should not be used as an alternate source directory because two manufacturers may use the same part number, by coincidence, for totally different devices.

## PART NUMBER GUIDE

The information in this guide allows you to break down each company's part numbering system into product temperature ranges, packaging variations, and functions. It is an invaluable tool for the elimination of costly and time-consuming ordering errors caused by lack of standardization from manufacturer to manufacturer in part numbering systems.

## APPLICATION NOTE DIRECTORY

Application note descriptions are arranged alphabetically by function and application category. Each note's description identifies the specific device or devices featured, provides a 25 to 30 word abstract, and identifies both the manufacturer that originated the note and the specific application note number. This section provides all the information necessary for you to update your application note files speedily, or thoroughly research the existence of application note material for a specific design problem.

## MILITARY PARTS DIRECTORY

Cross reference chart identifies all IC devices having received JAN qualification. This chart includes a cross reference listing of device numbers and corresponding military standard 38510 slash numbers and vice versa.

## MILITARY DEVICE TESTING TABLE

This table identifies IC manufacturers who test to military standard 38510 and the screening to military standard 883 that they provide.

## MILITARY PARTS INDEX

This guide to JAN qualified parts makes it possible to search devices by function, and to determine if a JAN qualified part exists for a particular functional need.

## ADVERTISERS' PRODUCT INDEX

This index directs you to detailed product information for the parts whose manufacturers have included data sheets in IC MASTER. When you are looking for data on a particular manufacturer's products, this index provides the fastest way to find the information you seek.

## ALTERNATE SOURCE DIRECTORY

The most comprehensive industry-wide, pin-for-pin, functional equivalent Alternate Source Directory ever compiled. This directory is updated by asking all IC manufacturers to identify each competitive device for which they make a pin-for-pin substitute.

## MASTER SELECTION GUIDE

Each guide is organized by specifications and categories to direct you easily and quickly to the device most likely to fill your requirements for a particular application. Once you find those devices that are closest to your needs, you see available sources, and are directed to additional data if provided by advertisers.

## MANUFACTURERS AND DISTRIBUTORS DIRECTORY

Locations and phone numbers are given for manufacturers' field sales offices, representatives, and distributors, both domestic and international.

## MASTER SELECTION GUIDE INDEX

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## PART NUMBER INDEX

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## PART NUMBER GUIDE

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## APPLICATION NOTE DIRECTORY

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# Typical Problems Solved by IC MASTER

- Find All Products That Meet Desired Specifications
- Obtain Data For A Particular Device • Decode Part Numbers
- Determine Alternate Sources • Plus Much More

An engineer can use IC MASTER to solve a wide variety of problems. Answers provided to engineers by IC MASTER can range from finding the device that best meets a particular set of specifications to helping to determine which family of devices should be used in building a system.

Some of the typical problems that IC MASTER can solve are illustrated in the following examples:

## Who makes a TTL 4-bit binary full-adder with look-ahead carry?

All functions are listed, in alphabetical order, in the Master Selection Guide Index. In this case, the engineer looks under adders; in the column adjacent to adders, he sees that all types of adders are listed. The particular adder being sought is covered in the Digital section of IC MASTER.

Now that the engineer knows that the devices he seeks are catalogued in the Digital section, he can turn to the Digital Master Selection Guide and see the page number where information on these devices can be obtained.

When he turns to this page, he will notice that certain device numbers and manufacturers are printed in bold face type while others appear in regular type face. Bold face type is used whenever a part's manufacturer has provided a data sheet for the device in IC MASTER. The page number assigned to the data sheet also appears in bold face type so that the engineer can turn to it directly.

## Who Makes a High-Speed 12-Bit, Analog-to-Digital Converter With Guaranteed $\pm 1/2$ LSB Linearity and 13- $\mu$ sec or Faster Conversion Time?

Many manufacturers make devices that meet these specifications including Analog Devices, Burr-Brown, Datal-Intersil, Data Device Corp., Harris, Hybrid Systems, Micro Networks, and Teledyne Philbrick.

The Master Selection Guide for Interface makes it possible for an engineer to find every device that meets the above specifications, regardless of who makes it, in seconds.

The Interface section is organized by product classification; an engineer can turn immediately to the category of interest such as analog switches with drivers, multiplexers, a/d converters with binary output a/d converters with decimal output, d/a converters, display drivers, error checking circuits, keyboard encoder-decoders, line drivers, line transceivers, memory and peripheral drivers, sense amplifiers, etc.

To find every 12-bit analog-to-digital converter with guaranteed  $\pm 1/2$  LSB linearity and 13- $\mu$ sec or faster conversion time, all an engineer has to do is turn directly to the analog-to-digital converter section of the Interface Master Selection Guide.

In this section, devices are organized by key parameters. Under resolution, the engineer finds 12-bit; next he looks under linearity error for  $\pm 1/2$  LSB and then he looks under conversion time for devices with 13- $\mu$ sec or faster specifications.

**MASTER SELECTION GUIDE-INDEX**

Function	Section	Function	Section
<b>A</b>		<b>Read</b>	Linear—Amplifiers, Special Purpose
AC Detector	Linear—Other Devices	RF Detector/Video Sense	Linear—Amplifiers, Special Purpose
Active Filter	Linear—Other Devices	Single Ended Input/Output	Linear—Consumer Circuits
Active Terminator	Digital—CMOS, Arithmetic Functions	Tone	Linear—Amplifiers, Special Purpose
Address	Digital—ECL, 10000, Arithmetic Functions	Transcon-Ductance	Linear—Amplifiers, Special Purpose
Address Latch Element	Digital—TTL, Arithmetic Functions	Video, IF and RF	Linear—Amplifiers, Special Purpose
Address Register	Microprocessors—Systems Components	Voltage Controlled	Linear—Amplifiers, Special Purpose
Address Selector	Microprocessors—Systems Components; 10000	Wideband	Linear—Amplifiers, Special Purpose
Alarm Circuits	Linear—Telecommunication Circuits	AM Radio Components	Linear—Consumer Circuits
ALU	Linear—Other Devices	AM/FM Radio Components	Linear—Consumer Circuits
Amplifier/Detector	Linear—Consumer Circuits	Analog Input	Linear—Consumer Circuits
FM IF Amplifiers	Linear—Other Devices	AM/FM Radio Components	Linear—Consumer Circuits
AG	Linear—Special Purpose	Analog Output	Linear—Other Devices
AM/FM IF	Linear—Consumer Circuits	Analog Shift Registers	Linear—Other Devices
AM/FM IF and AF	Linear—Consumer Circuits	Analog Signal Average	Linear—Other Devices
Audio	Linear—Amplifiers, Special Purpose	Analog Switches	Interface—Analog Switches
Audio, Power	Linear—Consumer Circuits	AND Gates	See Gates
CATV	Linear—Amplifiers, Special Purpose	Appliance Control Devices	Linear—Consumer Circuits
Current	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—CMOS, Arithmetic Functions
Differential/Cascade	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—CMOS, Arithmetic Functions
Differential Input/Output	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—ECL, Arithmetic Functions
Followers	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—MOS, Arithmetic Functions
Front End	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—TTL, Arithmetic Functions
Hearing Aid	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—TTL, Miscellaneous
Instrumentation	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—Special
Isolation	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—Special
Limiting	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—Special
Linear/Analog	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—Special
Log/Analog	Linear—Amplifiers, Special Purpose	Arithmetic Functions	Digital—Special
		Arithmetic Functions	Digital—Special

**IC MASTER**

**INTERFACE-Analog to Digital Converters (Cont'd)**

Device	Resolution	Linearity	Conversion Time	Manufacturer
AD7121	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7122	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7123	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7124	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7125	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7126	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7127	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7128	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7129	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7130	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7131	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7132	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7133	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7134	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7135	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7136	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7137	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7138	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7139	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7140	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7141	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7142	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7143	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7144	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7145	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7146	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7147	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7148	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7149	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7150	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7151	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7152	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7153	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7154	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7155	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7156	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7157	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7158	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7159	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7160	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7161	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7162	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7163	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7164	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7165	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7166	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7167	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7168	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7169	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7170	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7171	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7172	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7173	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7174	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7175	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7176	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7177	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7178	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7179	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7180	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7181	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7182	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7183	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7184	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7185	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7186	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7187	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7188	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7189	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7190	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7191	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7192	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7193	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7194	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7195	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7196	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7197	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7198	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7199	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices
AD7200	12-bit	$\pm 1/2$ LSB	13 $\mu$ sec	Analog Devices





# Ways to Use IC Master

## I need to choose between a full custom or a semi-custom/gate array solution to my design problem.

To help designers weigh custom solutions against semi-custom/gate array approaches, IC MASTER provides a Master Selection Guide on Custom/Semi-Custom and a special section entitled "Options for Going Custom." In this section the advantages and disadvantages of various custom/semi-custom techniques are covered.

The capabilities of IC manufacturers are tabulated for easy comparison; additional information such as available design aids and testing services is also provided.

## My application requires microcomputer boards. How do I start?

Single and multiple board microcomputers are arranged by manufacturer. Under each manufacturer, boards are grouped in sequence according to data word size and, within that grouping, according to the microprocessor on which they are based. Hardware and software support are listed for each board.

A supplementary selection guide is included for microcomputer support boards. The boards are grouped according to supported computer systems.

## With so many microprocessors available, where do I begin?

Simply turn to the Master Selection Guide for Microprocessors. There you will find a listing of all microprocessors currently available and key parameters allowing you to narrow down your selection to a range of products that will meet your major requirements.

Once the microprocessor that best fits the application has been chosen, the next step is to go to the "system components" section. Here all of the available peripheral devices that work with each microprocessor are arranged by function. Thus, if the microprocessor that has been selected is the 8048, system components specifically developed for use with the 8048 are listed, organized by function.

A "general purpose" section follows the "system components" section and describes devices that can be used with more than one microprocessor family.

Finally, hundreds of pages of the latest microprocessor data sheets, provided by IC manufacturers, are presented, arranged in alphabetical order by manufacturer. Each data sheet is easily found thanks to bold-faced page-number references in the Master Selection Guide.

### Options for Going Custom

For many applications, standard integrated circuits may be inappropriate from the standpoint of cost, size, power consumption or reliability. Moreover, unique features demanded by proprietary products often require entirely new circuit configurations. As a result, custom ICs are assuming an increasingly important role in system design.

Custom IC suppliers report that the chief benefits enjoyed by nearly all customer-circuit users are low-cost parts and cost savings resulting from reduced printed-circuit board space, parts handling, inventory, testing requirements and system maintenance. Obtaining these benefits requires careful selection of the manufacturer. The manufacturer's reputation and other approaches which ultimately effect economics.

Custom IC suppliers, in addition to standard and custom options for implementing new system designs include custom ICs, microcomputers, custom microprocessors, custom ICs, microcomputers and custom ICs, or a mixture of these. The system development strategy used depends upon marketing objectives and may require selection of standard or custom ICs, then with either a standard or custom IC, then with either a standard or custom IC, then with either a standard or custom IC, then with either a standard or custom IC.

Another option is to alter a standard microprocessor or custom IC, either by using a full or partial custom IC, or by using a full or partial custom IC. Customizing standard products can reduce development time and risks of a full custom IC, in some cases, semiconductor or custom circuits can be used. Customizing standard products can reduce development time and risks of a full custom IC, in some cases, semiconductor or custom circuits can be used.

All approaches require up-front decisions concerning development, production turnaround times, prototype and production quantities, interconnectability, testability, and user support/infrastructure, and the user's manufacturing capabilities. The cost factor, however, is cost.

The cost of a nonstandard IC includes design, testing, wafer and chip processing, and assembly.

**Design and Tooling:** Until recently, options for implementing new designs in custom ICs. When standard ICs had to be used, production had to be committed to great production quantities. Custom ICs have not been used in great quantities. However, custom ICs have been used in great quantities. However, custom ICs have not been used in great quantities.

**Full Custom:** In this approach, a collection of functions is performed on a chip. However, the user's manufacturing capabilities and equipment for chip area and testability are of the most efficient use of the available area. However, custom ICs have not been used in great quantities.

**Semi-custom:** This approach involves interconnecting a chip called a gate array or a gate array. The chip is processed just as if it were a standard IC. However, the user's manufacturing capabilities and equipment for chip area and testability are of the most efficient use of the available area.

Standard circuits are recommended for production volumes below 1000 units. Custom ICs are recommended for production volumes above 1000 units. Custom ICs are recommended for production volumes above 1000 units.

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# Ways to Use IC Master

### PART NUMBER INDEX

Part No.	Device	Manufacturer	Part No.	Device	Manufacturer	Part No.	Device	Manufacturer
6508	7400	TI	6508	7400	TI	6508	7400	TI
6508	7400	TI	6508	7400	TI	6508	7400	TI
6508	7400	TI	6508	7400	TI	6508	7400	TI

• IC MASTER 1983

### IC MASTER CROSS REFERENCE - COMMERCIAL TO MILITARY

Device	Manufacturer	Part No.	Device	Manufacturer	Part No.
6508	TI	6508	6508	TI	6508
6508	TI	6508	6508	TI	6508
6508	TI	6508	6508	TI	6508

• IC MASTER 1983

**If an engineer knows that the basic part number is 6508, where does he look first?**

The place to look is the part number index. Here, all of the prefixes and suffixes have been stripped away to leave only the basic number. This makes it possible to see the manufacturer of every part with the same base number at a glance. Parts with the same base number, it should be kept in mind, are not necessarily identical; in fact, one could be a memory while another might be a linear device. (To find replacements, one should look in the alternate source directory, not in the part number index.)

Under 6508, the engineer would see a number of devices listed and the page and line numbers where data is given for each of the devices. If an application note concerning any of the devices is available, the location of its listing is also shown.

**I design to military specs. Where can I find the latest OPL devices?**

The IC MASTER includes a comprehensive military parts directory. Within this directory, tables and charts are provided to answer virtually every information need of the engineer involved in military or high-reliability equipment design.

All integrated circuits with JAN qualification are listed in IC MASTER. A cross reference table, relating device numbers to mil spec numbers, is arranged according to device number. A second table, arranged by M-38510 number, makes it possible to look up the part number when the mil spec number is known.

In addition, ICs are also organized by product section and function, which allows the user to find the proper IC without having to know either the commercial or the military part number.

This section of IC MASTER also includes a table showing the capability of manufacturers to perform MIL-STD-883 screening and high-reliability testing.

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Systems 1902

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# INTRODUCTION TO INTERFACE

The Master Selection Guide provides sufficient information for making initial product selections. All devices that appear in this section, both in the initial selection guide and the data pages, are included in the indexes. These index listings lead to the page and line on that page where each device appears.

In the Interface Section the selection parameters differ drastically for each category; therefore each has its own format. The analog to digital converter category has two formats: one for binary output devices and another for decimal units. Some of the products in this section, primarily analog to digital and digital to analog converters, may be hybrids; the hybrids listed are only those packaged to be compatible with ICs.

This section is not complicated by reference to package styles; the package style suffixes are usually deleted. For more information on each companies' suffixes, see the Part Number Guide. Throughout the Master Selection Guide, each full military temperature range (-55°C to 125°C) device is indicated by a (†) before the manufacturer's name. Manufacturers' names are normally spelled out; however, a few are abbreviated and the abbreviations are explained on page 2809.

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Intersil Corporation	2988
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Micro Power Systems	3046
Mitel Semiconductor	3048
Motorola Semiconductor	3060
Plessey Semiconductors	3071
Siliconix	3077
Sprague Electric Co.	3087
Teledyne Semiconductor	3091
Texas Instruments	3103
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The manufacturers listed above have provided detailed information on their latest and most significant products.

# EINFÜHRUNG INTERFACE- SCHALTUNGEN

Der Master Selection Guide für Interfaceschaltungen enthält alle Informationen, die Sie für die Erstauswahl Ihres Produkts benötigen. Die Bauteile, die in diesem Abschnitt erscheinen, sowohl im Selection Guide als auch auf den Datenblättern, sind in allen Master Indexes enthalten. Diese Register verweisen auf die Seite und Zeile, auf der das entsprechende Bauelement vorkommt.

Im Interface-Teil unterscheiden sich die Auswahl-Parameter drastisch für jede Kategorie; daher hat jede ihre eigene Tabelle. Für Analog-Digital-Wandler gibt es zwei Tabellen: eine für Bauteile mit Binär-Ausgang und eine weitere für Dezimal-Bauteile. Einige Produkte in diesem Teil, hauptsächlich Analog-Digital und Digital-Analog Wandler, können hybride Bauelemente sein. Es werden nur solche hybriden Bauteile aufgeführt, deren Gehäuse kompatibel zu ICs sind.

Dieser Abschnitt wird nicht durch Hinweise auf Gehäuseformen kompliziert. Die entsprechenden Suffixe sind meistens weggelassen. Weitere Information über die Suffixe jedes Herstellers erhalten Sie über das Numerische Typenverzeichnis. Im ganzen Master Selection Guide sind alle Bauteile mit militärischem Temperaturbereich ( $-55^{\circ}\text{C}$  bis  $125^{\circ}\text{C}$ ) durch ein Kreuz (†) vor dem Namen des Herstellers gekennzeichnet. Die Namen der Hersteller sind normalerweise ausgeschreiben; einige jedoch sind abgekürzt. Die Abkürzungen werden auf S. 2809 erklärt.

# INTRODUCTION AUX INTERFACES

Le Guide Général de Sélection fournit suffisamment de renseignements pour permettre des sélections initiales de produits. Tous les appareils cités dans cette Section, à la fois dans la Premier Guide de Sélection et dans les feuilles de données, sont inclus dans tous les index. Ces index indiquent à quelle page et à quelle ligne sur cette page il a été fait mention de tel appareil.

Dans la Section "Interfaces", les paramètres de sélection changent pour chaque catégorie, ce qui implique un format spécial pour chacun. Les convertisseurs analogues et digitaux ont deux formats : un pour les appareils à sortie binaire et un autre pour les unités décimales. Certains produits de cette Section, notamment les convertisseurs d'analogie à digital et de digital à analogie, peuvent être des hybrides. Les hybrides énumérés sont ceux qui possèdent un boîtier compatible aux circuits intégrés.

Cette Section ne fait pas référence aux types de boîtier; les suffixes indiquant le type de boîtier sont généralement omis. Pour plus d'information sur les suffixes employés par chaque société, reportez-vous au Guide des Numéros de Pièces. Dans tout le Guide Général de Sélection chaque appareil avec sélection complète de températures imposées par l'Armée ( $-55^{\circ}\text{C}$  à  $125^{\circ}\text{C}$ ) est indiqué par le signe (†), juste avant le nom du fabricant, Les noms des fabricants sont généralement écrits en entier, certains cependant sont abrégés. Reportez-vous à la page 2809 pour connaître la signification de ces abréviations.



# INTRODUCCIÓN A ZONA INTERFACIAL

La Guía Maestra de Selección provee suficiente información para hacer selecciones iniciales de producto. Todas las componentes que aparecen en esta sección, ya sea en la guía de selección inicial o en las páginas de datos, están incluidas en todos los otros índices. Estas listas de índices los conduce a la página y línea de aquella página donde se encuentra cada componente.

En la sección de Zona Interfacial la selección de parámetros difiere drásticamente para cada categoría; de tal manera que cada una tiene su propio formato. La categoría de convertidores analógicos a digitales tiene dos formatos: uno para componentes de salida binaria y otro para unidades decimales. Algunos de los productos en esta sección, principalmente convertidores analógico a digital y digital a analógico, pueden ser híbridos; los híbridos en la lista son solo aquellos de estilo constructivo compatible con CI\*.

Esta sección no es complicada por referencias al estilo constructivo; los sufijos que denotan estilo constructivo han sido generalmente omitidos. Para información adicional sobre sufijos de las compañías, refiérase a la Guía de Número de Pieza. A lo largo de la Guía de Selección Maestra, cada intervalo completo de temperatura para uso militar (55°C a 125°C) de la componente aparece indicada por el signo (†) que precede al nombre del fabricante. Los nombres de los fabricantes no son generalmente deletreados; sin embargo, algunos aparecen abreviados y las abreviaturas son explicadas en la página 2809.

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INTERFACE-Analog Switches

Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line	Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line														
<b>Switches with Drivers</b>																													
SPST	CMOS	50	± 11	± 15,5	IH5140M	† Intersil (3028)	5	2xSPST	CMOS	75	± 11	± 15,5	IH200M	† Intersil (Cont'd)	65														
			± 15	± 15,5	DG5040A	† Silcoix (3077)							IH5041M	† Intersil (3021)															
					DG5040C	Silcoix (3077)							± 15	± 15,5		IH5041-2	† Harris (2882)												
	75	± 10	± 15,5	IH5140C	Intersil (3028)	80							± 10	± 15,5		IH200C	Intersil	70											
		± 11	± 15,5	HI5040-2	† Harris (2882)								± 14	± 15		IH5041C	Intersil (3021)												
				HI5040-5	† Harris (2882)								± 15	± 15		DG200C	Intersil (3031)												
			IH5040M	† Intersil (3021)	IH5200C	Intersil (3031)																							
	JFET	6	-5 to 10	± 15	± 15,5	CAG6							† TeledyneC	10		JFET	6	-6 to 10	-18,15	-18,15	CAG27	† TeledyneC	85						
				-5 to 4	± 15,5	CAG10A							† TeledyneC								10	-6 to 10		-18,15	CAG27-10	† TeledyneC			
		10	-5 to 10	± 15	CAG6-10	† TeledyneC							15								-7.5 to 15	± 15,5		DG180A	† Intersil	90			
-5 to 4			± 15,5	CAG10C	† TeledyneC	± 10	-18,12	DG180B	Intersil																				
15		-10 to 5	± 15,5	CAG10D	† TeledyneC	20			DG180A	† Silcoix (3077)	95																		
									DG180B	Silcoix (3077)																			
30		± 5	± 5,5	CAG10	† TeledyneC	25			DG141A	† Intersil	100																		
		± 8	-18,12	IH5001	Intersil					IH5005		† Intersil																	
50		-10 to 4	± 15,5	CAG10B	† TeledyneC	30			AH0141	† National	105																		
		-10 to 5	-15,5	CAG14	† TeledyneC					AH0141C		National																	
60	± 8	-18,12	IH5002	Intersil	35			DG141A	† Silcoix (3077)	110																			
	± 10	± 15	2110BE	† TeledyneC					DG141B		Silcoix (3077)																		
100	± 10	± 18,5	CAG30	† TeledyneC	40			DG151A	† Intersil	115																			
	± 5	± 15	2107BE	† TeledyneC					AH0151		† National																		
PMOS	80-300	± 10	-20,10	TL610I	TI	45	PMOS	80-300	± 10	-20,10	TL610M	† TI	50	JFET	10	-7.5 to 15	± 15,5	DG180A	† Intersil	120									
																						DG180B	Intersil						
2xSPST	CMOS	100-400	± 10	-20,10	TL610C	TI	55	2xSPST	CMOS	35	± 10	± 15,5	IH5048M	† Intersil (3021)	125														

DT<sup>1</sup> means four terminals with a pair of normally open and normally closed contacts.

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## INTERFACE-Analog Switches (Cont'd)

Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line	Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line
<b>Switches with Drivers</b>								<b>4xSPST Common Output PMOS</b>							
<b>(Cont'd)</b>								<b>(Cont'd)</b>							
2xSPST	JFET	50	-10 to 15 -12.5 to 10	± 15.5 -20, 10.5	AM281 DG181B IH181C DG181B	National Intersil Intersil Siliconix		4xSPST Var. Comb. of Normally On/Off, with Disable	FET	85	± 10	± 15	SW03B SW03F SW04B SW04F	† PMI PMI † PMI PMI	
			-15 to 10	-20, 10.5	AM281	National	5								65
			± 7.5	± 15	DG152A AH0152 AH0152C DG152A	† Intersil † National National † Siliconix (3077)		4xSPST Various Combinations of Normally On/Off	FET	80 *	± 10	± 15	SW-201B SW-202B	† PMI † PMI	
			± 8	-18, 12	DG133B	Intersil Siliconix (3077)	10			85	± 10	± 15	SW01B SW01F SW02B SW02F	† PMI PMI † PMI PMI	70
			± 10	-18, 12 -18, 15	IH5004 CAG13 CAG42 CAG45A	Intersil † TeledyneC † TeledyneC † TeledyneC				100	± 10	± 15	SW-201F SW-202F	PMI PMI	
			75	-7.5 to 15 -10 to 15	AM182 DG182A IH182M DG182A	† National † Intersil † Intersil † Siliconix (3077)	15	JFET	100 200	± 10 ± 10	-18, 15 ± 15, 0		CAG48A LF11201 LF11202 LF11331 LF11332 LF11333	† TeledyneC † National † National † National † National † National	75
			-12.5 to 10	-20, 10.5	AM182	† National	20			250	± 10	± 15, 0	LF13201 LF13202 LF13331 LF13332 LF13333	National National National National National	80
			-15 to 10	-20, 10.5	DG182A IH182M AM281 DG182A	† Intersil † Intersil National † Siliconix									
			± 10	± 15	SW-05B TL182M	† PMI † TI	25	PMOS	200-600	± 10	-20, 10, 5		AH0015 AH0015C	† National National	85
			80	± 8 ± 10	DG434A DG134A IH5007 AH0134 AH0134C DG134A	Intersil † Intersil Intersil † National National † Siliconix (3077)		4xSPST CMOS	50	± 15	± 15	HI201HS-2 HI201HS-5	† Harris (2868) Harris (2868)		
			-15 to 10	-20, 10.5	DG182B IH182C AM282 DG182B	Intersil Intersil National Siliconix (3077)	35			75	± 11	± 15, 5	IH201M IH202M IH5052M IH5053M	† Intersil † Intersil † Intersil † Intersil	90
			± 15		SW-05F SW-05G	PMI PMI				80	± 10	± 15	MP2010IA MP2010IB	† MicroPwr MicroPwr (3047)	
			100	-10 to 15	DG182B IH182C AM282 DG182B	Intersil Intersil National Siliconix (3077)	40						IH201C IH202C IH5052C IH5053C	Intersil Intersil Intersil Intersil	95
			-15 to 10	-20, 10.5	DG182B IH182C AM282 DG182B	Intersil Intersil National Siliconix					± 14	± 15	DG201M IH5201M	† Intersil † Intersil	100
			± 5.5	± 15	DG452A DG152B	Intersil Siliconix (3077)	45			90	± 15	± 15	AD7590DIB AD7590DIK AD7591DIB AD7591DIK DG221A DG221C	AD AD AD AD † Siliconix (3077) Siliconix (3077)	105
			± 8	-18, 12	DG134B DG134B	Intersil Siliconix (3077)					± 15	± 15	ADG201A HI201-2	† AD † Harris (2868)	
			± 10	± 15, 5	TL182C TL182I	TI TI	50						AD7590DIB AD7590DIK AD7591DIB AD7591DIK DG221A DG221C	AD AD AD AD † Siliconix (3077) Siliconix (3077)	110
			300	-7.5 to 15	DG281A DG281B	† Siliconix (3077) Siliconix (3077)					± 14	± 15	DG201C IH5201C	Intersil Intersil	
			PMOS	75-200 ± 10	DGM111A DGM111A	† Siliconix (3077) † Siliconix					± 15	± 15	AD7510DIJ AD7510DIK AD7510DIS AD7511DIJ AD7511DIK AD7511DIS AD7511DIT	AD AD † AD AD AD † AD † AD	115
			75-250	-5 to 10 ± 10	DGM111B DGM111B	Siliconix (3077) Siliconix							ADG201B ADG201C	AD AD	
			100-450	± 10	DG111 DG112	† Intersil † Intersil	55						ADG201C HI201-5 MP7510DIJ MP7510DIK	AD Harris (2868) MicroPwr (3047) MicroPwr (3047)	120
4xSPST Common Output PMOS		100-450	± 10	-20, 10 -20, 10, 5	DG116 DG118	† Intersil † Intersil									(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Analog Switches (Cont'd)

Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line	Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line
<b>Switches with Drivers</b>								<b>(Cont'd)</b>							
4xSPST	CMOS	100	± 15	± 15	(Cont'd)			4xSPST	CMOS	850	± 7.5	± 7.5	(Cont'd)		
					MP7510DIS	† MicroPwr (3047)	5	MN4016B	† Panasonic (828,834)				CD4016A	† RCA (839,840)	70
					MP7511DIJ	MicroPwr (3047)		CD4016AE	RCA (839,840)				CD4016B	† RCA (839,840)	
					MP7511DIK	MicroPwr (3047)		CD4016BE	† RCA (839,840)				CD40168E	† RCA (839,840)	
					MP7511DIS	† MicroPwr (3047)	5	HCC4016B	† SGS				HCF4016B	SGS	
					MP7511DIT	† MicroPwr (3047)		HEF4016	† Signetics (864)				CM4016A	† Solitron	
					DG308A	† Siliconix (3077)		CM4016AE	Solitron				CM4116A	† Solitron	75
					DG308C	Siliconix (3077)		CM4116AE	Solitron						
					DG309A	† Siliconix (3077)									
					DG309C	Siliconix (3077)									
		175	± 15	± 15	DG201A	† Siliconix (3077)	10						SD5001	Siliconix (3080)	
					DG201AA	† Siliconix (3077)							SD5002	Siliconix (3080)	
					DG202A	† Siliconix (3077)		DMOS	70	± 7.5	± 15		SD5000	Siliconix (3080)	80
					DG202B	Siliconix (3077)				± 10	± 15		SD5200	Siliconix (3080)	
					DG211C	Siliconix (3077)				± 15	± 15				
					DG212C	Siliconix (3077)	15								
		200	± 15	± 15	DG201AB	† Siliconix (3077)		JFET	10	± 10	- 18,2.8		CAM604A	† TeledyneC	
					DG201AC	† Siliconix (3077)			30	± 7.5	± 15,5		AM193	National	
					DG201ANSA	† Siliconix (3077)			50	- 10 to 5	- 15,5		CAG49	† TeledyneC	85
					DG201ANSB	Siliconix (3077)				± 10	± 18, - 7		CAG50	† TeledyneC	
					DG201ANSC	Siliconix (3077)			60	± 10	- 18,5		CAG49	† TeledyneC	
					DG201B	Siliconix (3077)	20		75	± 10	± 15		SW-7510A	† PMI	
					DG201C	Siliconix (3077)							SW-7510B	† PMI	
					DG202C	Siliconix (3077)							SW-7510E	PMI	
		280	± 7.5	± 7.5	F4066BC	Fairchild	25						SW-7510F	PMI	90
					F4066BM	† Fairchild							SW-7511A	† PMI	
					HD14066B	Hitachi							SW-7511B	† PMI	
					MC14066BA	† Motorola							SW-7511E	PMI	
					MC14066BC	Motorola							SW-7511F	PMI	
					MC54HC4066	† Motorola							AM194	National	95
					MC74HC4066	Motorola	30				± 15,5				
					CD4066BC	National		5xSPST Common Output							
					CD4066BM	† National			PMOS	100-450	± 10	- 20,10	DG123	† Intersil	
					MN4066B	† Panasonic (828)							DG123A	† Siliconix	
					MN74HC4066	† Panasonic (831)							DG125	† Intersil	
					CD4066A	† RCA (840)	35						DG125A	† Siliconix (3078)	
					CD4066AE	RCA (840)							DG123B	Intersil	100
					CD4066B	† RCA (840)			125-500	± 10	- 20,10		DG123B	Intersil	
					CD4066BE	RCA (840)							DG123B	Siliconix	
					CD74HC4066	RCA (843)							DG125B	Intersil	
					HCC4066B	† SGS	40						DG125B	Siliconix (3078)	
					HCF4066B	SGS									
					HEF4066B	Signetics (864,898)		SPDT for D/A							
					HEF4066B	Signetics (864)			NPN-PNP						
					HEF4066B	† Signetics (864)				10	0 to - 10	- 15	CDA1-3	† TeledyneC	
					N4066A	Signetics					10	10	CDA4A	† TeledyneC	105
					883/4066B	† SSS	45				± 10	- 15,5	CDA23	† TeledyneC	
					SCL4066B	SSS						± 15	CDA6	† TeledyneC	
					TC4066B	Toshiba		SPDT <sup>1</sup>	CMOS	35	± 10	± 15,5	IH5050M	† Intersil (3021)	
		400	± 7.5	± 7.5	HD14016B	Hitachi				45	± 10	± 15,5	IH5050C	† Intersil (3021)	
					MC14016BA	† Motorola				50	± 11	± 15,5	IH5142M	† Intersil (3028)	110
					MC14016BC	Motorola					± 15	± 15	HI301	Harris (2872)	
					883/4016B	† SSS							HI305	Harris (2872)	
					SCL4016B	SSS							HI387	Harris (2877)	
					CM4016A	† Solitron							DG301AA	† Siliconix (3078)	
					CM4016AE	Solitron							DG301AB	Siliconix (3078)	115
					TC4016B	Toshiba	55						DG301AC	Siliconix (3078)	
					F4016C	Fairchild							DG305AA	† Siliconix (3078)	
		800	± 7.5	± 7.5	F4016M	† Fairchild							DG305AB	Siliconix (3078)	
					883/4416B	† SSS							DG305AC	Siliconix (3078)	
					SCL4416B	SSS							DG387AA	† Siliconix (3078)	120
					MC54HC4016	† Motorola							DG387AB	Siliconix (3078)	
					MC54HC4316	† Motorola							DG387AC	Siliconix (3078)	
					MC74HC4016	Motorola									
					MC74HC4316	Motorola							HI5050-2	† Harris (2882)	
					CD4016BC	National							HI5050-5	Harris (2882)	
					CD4016BM	† National	65						DE5042A	† Siliconix (3078)	125
													DE5042C	Siliconix (3078)	
										75	± 10	± 15,5	HI5142C	† Intersil (3028)	
							(Continued)								(Continued)

DT<sup>1</sup> means four terminals with a pair of normally open and normally closed contacts.







**IC MASTER**

**INTERFACE—Analog Switches (Cont'd)**

Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line	Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line
<b>Switches with Drivers</b>								<b>(Cont'd)</b>							
2xDPST	CMOS	50	± 15	± 15				2xDPST	JFET	75	- 10 to 15	± 15,5			
								(Cont'd)							
								IH185M † Intersil							
								DG185A † Siliconix (3079)							
								- 12.5 to 10							
								- 20, 10, 5							
								AM185 † National							
								- 15 to 10							
								- 20, 10, 5							
								DG185A † Intersil							
								IH185M † Intersil							
								DG185A † Siliconix							
								DG426A Intersil							
								DG126A † Intersil							
								AH0126 † National							
								AH0126C National							
								DG126A † Siliconix (3079)							
								DG185B Intersil							
								IH185C Intersil							
								AM285 National							
								DG185B Siliconix (3079)							
								- 15 to 10							
								- 20, 10, 5							
								DG185B Intersil							
								IH185C Intersil							
								AM285 National							
								DG185B Siliconix							
								± 5.5 ± 15							
								DG154B Intersil							
								DG454A Intersil							
								DG154B Siliconix (3079)							
								± 8 - 18, 12							
								DG126B Siliconix (3079)							
								125 ± 10 ± 15,5							
								TL185M † TI							
								150 ± 10 ± 15,5							
								TL185C TI							
								TL185I TI							
								300 - 7.5 to 15 ± 15,5							
								DG284A † Siliconix (3079)							
								DG284B Siliconix (3079)							
								3xDPST Common Output							
								PMOS 100-450 ± 10 - 20, 10							
								DG120 † Intersil							
								- 20, 10, 5							
								DG121 † Intersil							
								DPDT <sup>1</sup> CMOS 50 ± 15 ± 15,5							
								IH5046A-2 † Harris (2882)							
								IH5046A-5 Harris (2882)							
								75 ± 11 ± 15,5							
								IH5046M † Intersil (3021)							
								± 15 ± 15,5							
								IH5046-2 † Harris (2882)							
								IH5046-5 Harris (2882)							
								80 ± 10 ± 15,5							
								IH5046C Intersil (3021)							
								JFET 10 ± 10 - 18, 12							
								DG145A † Intersil							
								AH0145 † National							
								AH0145C National							
								DG145A † Siliconix (3079)							
								15 ± 7.5 ± 15							
								DG163A † Intersil							
								AH0163 † National							
								AH0163C National							
								DG163A † Siliconix (3079)							
								± 8 - 18, 12							
								DG445A Intersil							
								DG145B Siliconix (3079)							
								20 ± 5.5 ± 15							
								DG163B Intersil							
								DG463A Intersil							
								DG163B Siliconix (3079)							
								30 ± 10 - 18, 12							
								DG139A † Intersil							
								AH0139 † National							
								AH0139C National							
								DG139A † Siliconix (3079)							
								35 ± 8 - 18, 12							
								DG439A Intersil							
								50 ± 7.5 ± 15							
								DG164A † Intersil							
								AH0164 † National							
								AH0164C National							
								DG164A † Siliconix (3079)							
								± 8 - 18, 12							
								DG139B Siliconix (3079)							
								80 ± 8 - 18, 12							
								DG442A Intersil							
								DG142A † Intersil							
								(Continued)							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Analog Switches (Cont'd)

Switch Function	Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line	Switch Function	Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line
<b>Switches with Drivers</b>								<b>(Cont'd)</b>							
DPDT	JFET	80	± 10	- 18, 12	AH0142 AH0142C DG142A	† National National † Silliconix (3079)	(Cont'd)	3xSPST Common Output	JFET	100	0 to (Drive-4)		IH5029C IH5029M	Intersil † Intersil	60
					DG164B DG464A DG164B	Intersil Intersil Silliconix (3079)	5				± 0.2		IH5013C IH5013M	Intersil † Intersil	
		100	± 5.5	± 15	DG142B	Silliconix (3079)				150	0 to (Drive-4)		IH5030C IH5030M	Intersil † Intersil	65
					DG142B	Silliconix (3079)					± 0.2		IH5014C IH5014M	Intersil † Intersil	
DPDT	PMOS	200-600	± 10	- 20, 10, 5	AH0014 AH0014C	† National National		3xSPST	JFET	100	0 to (Drive-4)		IH5031C IH5031M	Intersil † Intersil	70
2xDPDT	PMOS	150-1000	0 to -18	-(5-21)	TDA1195	Siemens	10				± 0.2		IH5015C IH5015M	Intersil † Intersil	
DP4T with Followers					TDA1029	Signetics				150	0 to (Drive-4)		IH5032C IH5032M	Intersil † Intersil	75
4xSPST Var. Comb. of Normally On/Off, with Disable	FET	80	± 10	± 15	SW-06B SW-06F	† PMI PMI					± 0.2		IH5016C IH5016M	Intersil † Intersil	
		100	± 10	± 15	SW-06G	PMI		4xSPST Common Output	JFET	—			G129 G130 G131 G132 G1350 G1360	Intersil † Intersil † Intersil † Intersil Intersil Intersil	80
		150	± 10	± 15	SW-201B SW-202B	† PMI † PMI	15						IH5025C IH5025M	Intersil † Intersil	85
		80	± 10	± 15	SW-201F SW-202F	PMI PMI				100	0 to (Drive-4)		IH5009C IH5009M AH5009C AM9709C AM97C09	Intersil † Intersil National National National	90
		100	± 10	± 15	SW-201G	PMI					± 0.2		IH5026C IH5026M	Intersil † Intersil	95
4PDT for D/A	PNP	10	± 10	± 15, 5	CDA28A	† TeledyneC	20						IH5010C IH5010M AH5010C AM9710C AM97C10C	Intersil † Intersil National National National	100
4PST	CMOS	50	± 15	± 15, 5	IH5047A-2 IH5047A-5	† Harris (2882) Harris (2882)							G124 MM451 MM551	Intersil † Intersil Intersil	105
		75	± 11	± 15, 5	IH5047M	† Intersil (3021)	25						G123 G123A G123B	Intersil † Silliconix (3080) Silliconix (3080)	110
			± 15	± 15, 5	IH5047-2 IH5047-5	† Harris (2882) Harris (2882)							G125 G126 G127 G128 G1330 G1340	† Intersil † Intersil † Intersil † Intersil Intersil Intersil	115
		80	± 10	± 15, 5	IH5047C	Intersil (3021)							IH5027C IH5027M	Intersil † Intersil	120
<b>Switches without Drivers</b>															
SPST	JFET	30	± 10		AM1000 AM1001	† National † National		4xSPST Two Outputs	JFET	—			G125 G126 G127 G128 G1330 G1340	† Intersil † Intersil † Intersil † Intersil Intersil Intersil	105
		50	± 15		IH5037C IH5037M	Intersil † Intersil	30						IH5028C IH5028M	Intersil † Intersil	110
		100	0 to (Drive-4)		IH5021C IH5021M IH5023C IH5023M	† Intersil † Intersil Intersil † Intersil							IH5011C IH5011M AH5011C AM9711C AM97C11C	Intersil † Intersil National National National	115
					AM1002	† National	35						IH5012C IH5012M	Intersil † Intersil	120
		150	0 to (Drive-4)		IH5038C IH5038M	Intersil † Intersil				100	0 to (Drive-4)		IH5028C IH5028M	Intersil † Intersil	120
					IH5022C IH5022M IH5024C IH5024M	Intersil † Intersil Intersil † Intersil	40						IH5012C IH5012M	Intersil † Intersil	
					IH5038C IH5038M	Intersil † Intersil					± 0.2		IH5012C IH5012M	Intersil † Intersil	
2xSPST Common Output	JFET	100	0 to (Drive-4)		IH5035C IH5035M	Intersil † Intersil	45						IH5012C IH5012M	Intersil † Intersil	
					IH5019C IH5019M	Intersil † Intersil							IH5012C IH5012M	Intersil † Intersil	
		150	0 to (Drive-4)		IH5034C IH5034M	Intersil † Intersil							IH5012C IH5012M	Intersil † Intersil	
					IH5018C IH5018M	Intersil † Intersil							IH5012C IH5012M	Intersil † Intersil	
2xSPST	JFET	100	0 to (Drive-4)		IH5033C IH5033M	Intersil † Intersil	50			150	0 to (Drive-4)		IH5028C IH5028M	Intersil † Intersil	115
					IH5017C IH5017M	Intersil † Intersil							IH5012C IH5012M	Intersil † Intersil	
		150	0 to (Drive-4)		IH5036C IH5036M	Intersil † Intersil	55						IH5012C IH5012M	Intersil † Intersil	
					IH5020C IH5020M	Intersil † Intersil					± 0.2		IH5012C IH5012M	Intersil † Intersil	

DT means four terminals with a pair of normally open and normally closed contacts.

# IC MASTER

## INTERFACE—Analog Switches (Cont'd)

Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line	Function	Switch Type	ON Resistance 25°C, Ω	Analog Signal Range V	Supply Voltage, V	Device	Source	Line
<b>Switches without Drivers</b>								<b>Multiplexers</b>							
<b>4xSPST</b>								<b>4 channel. See Switches 4xSPST</b>							
PMOS 200-600								Crosspoint Switches							
MM452 † Intersil (Cont'd)								MC142100 Motorola (3536)							
MM552 Intersil								MC145100 Motorola (3536)							
5xSPST Common Output plus Output Switch								MC3416 Motorola (3060)							
PMOS —								RC4444 Raytheon							
G117A † Intersil								RM4444 † Raytheon							
G117B Intersil								CD22100 RCA (840)							
G117A † Siliconix								CD22101 RCA (840)							
5xSPST Common Output								CD22102 RCA (840)							
PMOS —								EFB7310 Thomson-CSF							
G116A † Intersil								4 Channel Differential							
G116B Intersil								CMOS 230 ± 15 ± 15							
G116A † Siliconix								MP7502S † MicroPwr (3047)							
G116B Siliconix								MP7509S † MicroPwr (3047)							
6xSPST Common Output								250 ± 15 ± 15							
PMOS —								MVD-409 Datal (2864)							
G115A † Intersil								MVD-409M † Datal (2864)							
G115B Intersil								270 ± 7.5 ± 7.5							
G118A † Intersil								MC14529BA † Motorola							
G118B Intersil								MC14529BC Motorola							
AM2009 † National								CD4529BC National							
AM2009C National								CD4529BM † National							
G115A † Siliconix (3080)								280 ± 7.5 ± 7.5							
G115B Siliconix (3080)								F4052BC Fairchild							
G118A † Siliconix (3080)								F4052BM † Fairchild							
2xDPST Common Output								HD14052B Hitachi							
PMOS —								MC14052BA † Motorola							
MM450 † Intersil								MC14052BC Motorola							
MM550 Intersil								MC54HC4052 † Motorola							
G122A † Siliconix (3080)								MC54HC4352 † Motorola							
3xDPST Common Output								MC74HC4052 Motorola							
PMOS —								MC74HC4352 Motorola							
G119A † Intersil								CD4052BC National							
G119B Intersil								CD4052BM † National							
G119A † Siliconix (3080)								MM4052B † Panasonic (828)							
G119B Siliconix (3080)								MM74HC4052 Panasonic (831)							
<b>Drivers</b>								CB4052B † RCA (839)							
High Current Switch (to drive power transistor switches)								CD4052BE RCA (839)							
SG1629 † SiliconG								CB74HC4052 RCA (843)							
SG3629 SiliconG								HCC4052B † SGS							
Pulse Width Modulator, Switch Driver								HCF4052B SGS							
PWM125 Siliconix (3081)								883/4052B † SSS							
PWM127 Siliconix (3081)								SCL4052B SSS							
PWM25 Siliconix (3081)								CM4052B † Solitron							
PWM27 Siliconix (3081)								CM4052BE Solitron							
2 Channel								TC4052B Toshiba							
D112C Intersil								300 ± 15 ± 15							
D112M † Intersil								AD7502J AD (2841)							
D113C Intersil								AD7502K AD (2841)							
D113M † Intersil								AD7502S † AD (2841)							
D120C Intersil								HI509-2 † Harris (2893)							
D120M † Intersil								HI509-5 Harris (2893)							
D121C Intersil								IH6208M † Intersil (3039)							
D121M † Intersil								320 ± 15 ± 15							
D139A † Siliconix								MP7502J MicroPwr (3047)							
D139B Siliconix								MP7502K MicroPwr (3047)							
D139C Siliconix								MP7509J MicroPwr (3047)							
4 Channel								MP7509K MicroPwr (3047)							
D129 Intersil								350 ± 15 ± 15							
D129A † Siliconix (3080)								IH6208C Intersil (3039)							
D129B Siliconix (3080)								400 ± 15 ± 15							
D469 Siliconix (3080)								HI1828A-2 † Harris (2903)							
SI7250 Siliconix (3080)								HI1828A-5 Harris (2903)							
6 Channel								D6506AA † Siliconix (3082)							
D123C Intersil								D6509AA † Siliconix (3082)							
D123M † Intersil								450 ± 15 ± 15							
D125C Intersil								D6506AB Siliconix (3082)							
D125M † Intersil								D6506AC Siliconix (3082)							
D123A † Siliconix								D6509AB Siliconix (3082)							
D123B Siliconix								D6509AC Siliconix (3082)							
D125A † Siliconix (3080)								750 ± 15 ± 15							
D125B Siliconix (3080)								MPC801 Burr-Brown (2850)							
CDR125A † TeledyneC								900 ± 10 ± 15							
								HI539-2 † Harris (2917)							
								HI539-5 Harris (2917)							
								1000 ± 13 ± 15							
								IH5208M † Intersil (3032)							
								1200 ± 13 ± 15							
								IH5208C Intersil (3032)							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

(Continued)

INTERFACE-Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Conver-sion Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Int-e-gra-ting	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
8	1/2	1	400	x											x		TDC1002 TRW-LSI (3704)	(Cont'd)
			1250	x	x			x							x	x	ADC-825MC Datal (2860)	
				x	x			x							x	x	ADC-825MM † Datal (2860)	
						x	x	x					x	x	x		MN5825 † MicroNet (3043)	
		1*	—	x				x					x				AM6108 AMD (3303)	5
			1000	x				x					x	x			AM6148 AMD (3303)	
		1.2	1800			x	x						x				HAS-0802 AD (2839)	
		1.5	1550	x		x	x						x	x	x		MN5100 MicroNet (3043)	
				x		x	x						x	x	x		MN5100H † MicroNet (3043)	
		2.5	35*	x							x				x		ADC0820B National	10
			650	x	x			x					x	x			ADC541B-8 † HybridSys	
				x	x			x					x	x			ADC541C-8 HybridSys	
						x	x	x					x	x	x		ADC542B-8 † HybridSys	
						x	x	x					x	x	x		ADC542C-8 HybridSys	
			915			x							x	x	x		MN5140 MicroNet (3043)	15
						x							x	x	x		MN5140H † MicroNet (3043)	
					x								x	x	x		MN5141 MicroNet (3043)	
					x								x	x	x		MN5141H † MicroNet (3043)	
					x								x	x	x		MN5142 MicroNet (3043)	
					x								x	x	x		MN5142H † MicroNet (3043)	20
				x									x	x	x		MN5143 MicroNet (3043)	
				x									x	x	x		MN5143H † MicroNet (3043)	
			1000			x							x	x	x		MN5130 MicroNet (3043)	
						x							x	x	x		MN5130H † MicroNet (3043)	
					x								x	x	x		MN5131 MicroNet (3043)	25
					x								x	x	x		MN5131H † MicroNet (3043)	
					x								x	x	x		MN5132 MicroNet (3043)	
					x								x	x	x		MN5132H † MicroNet (3043)	
				x									x	x	x		MN5133 MicroNet (3043)	
				x									x	x	x		MN5133H † MicroNet (3043)	30
				x	x	x							x	x	x		MN5150 MicroNet (3043)	
		2.8	650			x	x	x					x	x	x		HSADC82 HybridSys	
			1000*			x	x	x					x	x	x		ADC83A Burr-Brown	
		6	1000			x							x	x	x		MN5120 MicroNet (3043)	
						x							x	x	x		MN5120H † MicroNet (3043)	35
					x								x	x	x		MN5121 MicroNet (3043)	
					x								x	x	x		MN5121H † MicroNet (3043)	
					x								x	x	x		MN5122 MicroNet (3043)	
					x								x	x	x		MN5122H † MicroNet (3043)	
				x									x	x	x		MN5123 MicroNet (3043)	40
				x									x	x	x		MN5123H † MicroNet (3043)	
		8	—	x									x				ZN437E Ferranti	
				x	x								x				ZN437J † Ferranti	
		9	4	x						x	x				x		TLC540 TI (541.3103)	
			125*	x	x								x	x			ZN448E Ferranti	45
				x	x								x	x			ZN448E-8E Ferranti	
				x	x								x	x			ZN448J † Ferranti	
			15	30	x	x		x					x				AD7574B AD (2838)	
				x	x			x					x				AD7574K AD (2838)	
				x	x			x					x				AD7574T † AD (2838)	50
				x	x			x					x				MP7574B MicroPwr (3046)	
				x	x			x					x				MP7574K MicroPwr (3046)	
					x			x					x				MP7574T MicroPwr (3046)	
			125*	x	x								x	x			ZN427E-8 Ferranti	
				x	x								x	x			ZN427J-8 † Ferranti	55

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Conver-sion Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-grating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
8	1/2																	
	17	300		x										x		NE5034	Signetics (3609)	(Cont'd)
	30	—		x									x			AD673J	AD (2838)	
				x									x			AD673S	†AD (2838)	
	40	5 *		x	x									x	x	AD7570J	AD	
				x	x									x	x	MP7570J	MicroPwr (3046)	5
		10 *		x						x				x		ADC0844B	National (3537)	
		175 *		x	x								x	x		AD570J	AD (2838)	
				x	x								x	x		AD570S	†AD (2838)	
				x	x								x	x		MCE570J	MCE	
				x	x								x	x		MCE570S	†MCE	10
	60MHz	—		x	x		x							x		TDC1025	†TRW-LSI (3704,3705)	
	66.6	—		x												AD8581L	AD	
	80	10		x											x	ADC0831B	National	
				x											x	ADC0832B	National	
				x											x	ADC0833B	National	15
				x											x	ADC0834B	National	
				x											x	ADC0838B	National	
	100	15		x						x				x		ADC0808	National	
				x						x				x		ADC0816C	National	
		30		x						x				x		ADC0816	†National	20
				x						x				x		ADC0808	TI	
		70				x							x	x	x	ADC-830	Datel (2863)	
							x						x	x	x	MN5065	MicroNet (3043)	
							x						x	x	x	MN5065H	†MicroNet (3043)	
						x							x	x	x	MN5066	MicroNet (3043)	25
						x							x	x	x	MN5066H	†MicroNet (3043)	
		75								x					x	MB4052	Fujitsu	
	100 *	875		x										x		ADC0802	†Intersil (2993)	
				x										x		ADC0802C	Intersil (2993)	
				x										x		ADC0803	†Intersil (2993)	30
				x										x		ADC0803C	Intersil (2993)	
				x										x		ADC0802	†National	
				x										x		ADC0802C	National	
				x										x		ADC0803C	National	
	108 *	6.8		x						x				x		MK50808	Mostek	35
				x						x				x		MK50816	Mostek	
	300	15 *		x						x						TL530	TI	
				x												TL532	TI	
	1250 *	20 *		x	x									x		4140	TeledyneP	
				x	x									x		4143	TeledyneP	40
				x	x									x		4143-01	†TeledyneP	
	1800	20		x	x							x		x		ADC-EK8B	Datel (2862)	
		25		x								x		x		TSC8700	TeledyneS	
				x								x		x		TSC8703	TeledyneS	
		43		x										x		ADC-ET88M	†Datel (2862)	45
		50		x										x		ADC-ET88C	Datel (2862)	
	3/4	15	30	x	x		x						x			AD7574A	AD (2838)	
				x	x		x						x			AD7574J	AD (2838)	
				x	x		x						x			AD7574S	†AD (2838)	
				x	x		x						x			MP7574A	†MicroPwr (3046)	50
				x	x		x						x			MP7574J	MicroPwr (3046)	
				x	x		x						x			MP7574S	†MicroPwr (3046)	

Bin.—Binary Off.—Offset Compl.—Complementary Magn.—Magnitude CTC—Compl. 2's Compl. Int. Ref.—Internal Reference Mux. In.—Multiplexed Inputs S&H—Sample and Hold Par. Out.—Parallel Output Ser. Out.—Serial Output

**IC MASTER**

**INTERFACE—Analog to Digital Converters (Cont'd)**

Bits Res.	Linear-ity Error ± LSB	Conver-sion Time ± ½ LSB μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-grating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
8	3/4	66.6	—	x												<b>AD7581K</b>	<b>AD (2838)</b>	(Cont'd)
		70	2.5 *	x						x				x		<b>SI520</b>	<b>Siliconix (3083)</b>	
				x						x				x		TL520	TI	
1	1.5	75		x							x			x		ADC0820C	National	
		2.5	1010		x										x	H55131B	† HybridSys	5
					x										x	H55131C	HybridSys	
	9	—			x	x							x	x		ZN449E-8	DDC	
					x	x							x	x		ZN449J-8	† DDC	
		125		x	x								x	x		<b>ADC-847A</b>	<b>Datal (2863)</b>	
		125 *		x	x								x	x		ZN449E	Ferranti	10
				x	x								x	x		ZN449J	† Ferranti	
	18	4		x						x	x				x	<b>TLC541</b>	<b>TI (541,3103)</b>	
	32	12.5		x						x					x	ADC0831C	National	
				x						x					x	ADC0832C	National	
				x						x					x	ADC0833C	National	15
				x						x					x	ADC0834C	National	
				x						x					x	ADC0838C	National	
	40	12.5		x						x				x		ADC0844B	National	
	100	2.5 *		x						x				x		TL521	TI	
		15		x						x				x		ADC0809C	National	20
				x						x				x		ADC0817	National	
	100 *	12.5		x										x		<b>ADC0804</b>	<b>† Intersil (2993)</b>	
				x										x		<b>ADC0804C</b>	<b>Intersil (2993)</b>	
				x										x		ADC0804C	National	
	110	12.5		x										x		ADC0805C	National	25
				x										x		ADC0804C	TI	
	112 *	15		x						x					x	μPD7001	NEC	
	300	15 *		x						x						TL531	TI	
				x												TL533	TI	
	2000	6.5 *		x									x		x	TC5091	Toshiba	30
		300		x									x		x	TC5090	Toshiba	
	2	50	255	x										x		ADC0800P	† National	
				x										x		ADC0800PC	National	
	300	25								x	x	x			x	MB4053	Fujitsu	
8 (analog input for microprocessors)																		
	1/2	2.5	1350	x						x						<b>MP20</b>	<b>Burr-Brown (2849)</b>	35
				x						x						<b>MP21</b>	<b>Burr-Brown (2849)</b>	
	32	50 *		x												MC14444	Motorola	
	400	1000		x						x		x				AD7583K	AD	
8 (analog to pulse width converter for microprocessor systems)																		
	1/2	—	15 *							x						MC14443	Motorola	
		32	50 *							x						MC14447	Motorola	40
			75 *							x						μA9708C	Fairchild	
										x						μA9708M	† Fairchild	
	350	25 *		x						x	x					MB4053	Fujitsu	
8 (D/A, A/D, with counter)																		
	1/2	500	150	x	x								x	x		<b>ADC-MC88C</b>	<b>Datal (2862)</b>	
				x	x								x	x		<b>ADC-MC88M</b>	<b>† Datal (2862)</b>	45
8 (video digital converter)																		
	1/2	0.05	1320	x										x		TDC1048	TRW-LSI	
9	1/2	1	500 *	x	x								x	x	x	ZN433BJ-9	† DDC	
				x	x								x	x	x	ZN433CJ-9	† DDC	
				x	x								x	x	x	ZN433J-9	† DDC	
9 (video A/D converter)																		
	1/2	0.02	2500	x					x							<b>TDC1019J</b>	<b>TRW-LSI (3704)</b>	50

† Military Temperature Range (−55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error ± LSB	Conver-sion Time μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-grating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																<b>(Cont'd)</b>		
8	1/2																	
		17	300	x										x		NE5034	Signotics (3609)	(Cont'd)
		30	—	x									x			AD673J	AD (2838)	
				x									x			AD673S	†AD (2838)	
		40	5 *	x	x									x	x	AD7570J	AD	
				x	x									x	x	MP7570J	MicroPwr (3046)	5
			10 *	x						x				x		ADC0844B	National (3537)	
			175 *	x	x								x	x		AD570J	AD (2838)	
				x	x								x	x		AD570S	†AD (2838)	
				x	x								x	x		MCE570J	MCE	
				x	x								x	x		MCE570S	†MCE	10
		60MHz	—	x	x		x								x	TDC1025	†TRW-LSI (3704.3705)	
		66.6	—	x												AD8581L	AD	
		80	10	x											x	ADC0831B	National	
				x											x	ADC0832B	National	
				x											x	ADC0833B	National	15
				x											x	ADC0834B	National	
				x											x	ADC0838B	National	
		100	15	x						x				x		ADC0808	National	
				x						x				x		ADC0816C	National	
			30	x						x				x		ADC0816	†National	20
				x						x				x		ADC0808	TI	
			70			x							x	x	x	ADC-830	Datel (2863)	
							x						x	x	x	MN5065	MicroNet (3043)	
							x						x	x	x	MN5065H	†MicroNet (3043)	
						x							x	x	x	MN5066	MicroNet (3043)	25
						x							x	x	x	MN5066H	†MicroNet (3043)	
			75							x					x	MB4052	Fujitsu	
		100 *	875	x										x		ADC0802	†Intersil (2993)	
				x										x		ADC0802C	Intersil (2993)	
				x										x		ADC0803	†Intersil (2993)	30
				x										x		ADC0803C	Intersil (2993)	
				x										x		ADC0802	†National	
				x										x		ADC0802C	National	
				x										x		ADC0803C	National	
		108 *	6.8	x						x				x		MK50808	Mostek	35
				x						x				x		MK50816	Mostek	
		300	15 *	x						x						TL530	TI	
				x												TL532	TI	
		1250 *	20 *	x	x									x		4140	TeledyneP	
				x	x									x		4143	TeledyneP	40
				x	x									x		4143-01	†TeledyneP	
		1800	20	x	x							x		x		ADC-EK8B	Datel (2862)	
			25	x								x		x		TSC8700	TeledyneS	
				x								x		x		TSC8703	TeledyneS	
			43	x										x		ADC-ET88B	†Datel (2862)	45
			50	x										x		ADC-ET88C	Datel (2862)	
		3/4	15	30	x	x		x					x			AD7574A	AD (2838)	
					x	x		x					x			AD7574J	AD (2838)	
					x	x		x					x			AD7574S	†AD (2838)	
					x	x		x					x			MP7574A	†MicroPwr (3046)	50
					x	x		x					x			MP7574J	MicroPwr (3046)	
					x	x		x					x			MP7574S	†MicroPwr (3046)	

Bin.—Binary Off.—Offset Compl.—Complementary Magn.—Magnitude CTC—Compl. 2's Compl. Int. Ref.—Internal Reference Mux. In—Multiplexed Inputs S&H—Sample and Hold Par. Out—Parallel Output Ser. Out—Serial Output



**IC MASTER**

**INTERFACE-Analog to Digital Converters (Cont'd)**

Bits Res.	Linearity Error ± LSB	Conversion Time ± 1/2 LSB μs	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Integrating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
8	3/4	66.6	—	x												<b>AD7581K</b>	<b>AD</b>	(2838)
		70	2.5 *	x						x				x		<b>SI520</b>	<b>Siliconix</b>	(3083)
				x						x				x		<b>TL520</b>	<b>TI</b>	
1	1.5	75		x							x			x		<b>ADC0820C</b>	<b>National</b>	
	2.5	1010			x										x	<b>HS5131B</b>	<b>† HybridSys</b>	5
					x										x	<b>HS5131C</b>	<b>HybridSys</b>	
	9	—			x	x							x	x		<b>ZN449E-8</b>	<b>DDC</b>	
					x	x							x	x		<b>ZN449J-8</b>	<b>† DDC</b>	
	125			x	x								x	x		<b>ADC-847A</b>	<b>Datel</b>	(2863)
	125 *			x	x								x	x		<b>ZN449E</b>	<b>Ferranti</b>	10
				x	x								x	x		<b>ZN449J</b>	<b>† Ferranti</b>	
	18	4		x						x	x				x	<b>TLC541</b>	<b>TI</b>	(541,3103)
	32	12.5		x						x					x	<b>ADC0831C</b>	<b>National</b>	
				x						x					x	<b>ADC0832C</b>	<b>National</b>	
				x						x					x	<b>ADC0833C</b>	<b>National</b>	15
				x						x					x	<b>ADC0834C</b>	<b>National</b>	
				x						x					x	<b>ADC0838C</b>	<b>National</b>	
	40	12.5		x						x				x		<b>ADC0844B</b>	<b>National</b>	
	100	2.5 *		x						x				x		<b>TL521</b>	<b>TI</b>	
		15		x						x				x		<b>ADC0809C</b>	<b>National</b>	20
				x						x				x		<b>ADC0817</b>	<b>National</b>	
	100 *	12.5		x						x				x		<b>ADC0804</b>	<b>† Intersil</b>	(2993)
				x						x				x		<b>ADC0804C</b>	<b>Intersil</b>	(2993)
				x						x				x		<b>ADC0804C</b>	<b>National</b>	
	110	12.5		x						x				x		<b>ADC0805C</b>	<b>National</b>	25
				x						x				x		<b>ADC0804C</b>	<b>TI</b>	
	112 *	15		x						x				x		<b>μPD7001</b>	<b>NEC</b>	
	300	15 *		x						x						<b>TL531</b>	<b>TI</b>	
				x						x						<b>TL533</b>	<b>TI</b>	
	2000	6.5 *		x									x		x	<b>TC5091</b>	<b>Toshiba</b>	30
		300		x									x		x	<b>TC5090</b>	<b>Toshiba</b>	
	2	50	255	x										x		<b>ADC0800P</b>	<b>† National</b>	
				x										x		<b>ADC0800PC</b>	<b>National</b>	
	300	25								x	x	x			x	<b>MB4053</b>	<b>Fujitsu</b>	
<b>8 (analog input for microprocessors)</b>																		
	1/2	2.5	1350	x						x						<b>MP20</b>	<b>Burr-Brown</b>	(2849)
				x						x						<b>MP21</b>	<b>Burr-Brown</b>	(2849)
	32	50 *		x												<b>MC14444</b>	<b>Motorola</b>	
	400	1000		x						x		x				<b>AD7583K</b>	<b>AD</b>	
<b>8 (analog to pulse width converter for microprocessor systems)</b>																		
	1/2	—	15 *							x						<b>MC14443</b>	<b>Motorola</b>	
	32	50 *								x						<b>MC14447</b>	<b>Motorola</b>	40
		75 *								x						<b>μA9708C</b>	<b>Fairchild</b>	
										x						<b>μA9708M</b>	<b>† Fairchild</b>	
	350	25 *		x						x	x					<b>MB4053</b>	<b>Fujitsu</b>	
<b>8 (D/A, A/D, with counter)</b>																		
	1/2	500	150	x	x								x	x		<b>ADC-MC88C</b>	<b>Datel</b>	(2862)
				x	x								x	x		<b>ADC-MC88M</b>	<b>† Datel</b>	(2862)
<b>8 (video digital converter)</b>																		
	1/2	0.05	1320	x										x		<b>TDC1048</b>	<b>TRW-LSI</b>	
9	1/2	1	500 *	x	x								x	x	x	<b>ZN433BJ-9</b>	<b>† DDC</b>	
				x	x								x	x	x	<b>ZN433CJ-9</b>	<b>† DDC</b>	
				x	x								x	x	x	<b>ZN433J-9</b>	<b>† DDC</b>	
<b>9 (video A/D converter)</b>																		
	1/2	0.02	2500	x				x								<b>TDC1019J</b>	<b>TRW-LSI</b>	(3704)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Conver-sion Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Int-egra-ting	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
10	—	3	—										x			MAS-1202 AD	(2839)	
	1/2	0.8	3600	x	x			x					x	x	x	ADC-816MC Datel	(2860)	
				x	x			x					x	x	x	ADC-816MM †Datel	(2860)	
		1	500 *	x	x								x	x	x	ADC-856C Datel	(2863)	5
				x	x								x	x	x	ADC-856M †Datel	(2863)	
				x	x								x	x	x	ZN433BJ-10 Ferranti		
				x	x								x	x	x	ZN433CJ-10 Ferranti		
				x	x								x	x	x	ZN433J-10 †Ferranti		
				x	x								x	x	x	ZN433CK-10 DDC		10
				x	x								x	x	x	ZN433K-10 DDC		
		1.4	1800 *			x	x						x			HAS-1002 AD	(2839)	
			3600	x	x			x					x	x	x	ADC-826MC Datel	(2860)	
				x	x			x					x	x	x	ADC-826MM †Datel	(2860)	
		1.8	755 *	x	x								x	x	x	AD579B AD	(2839)	15
				x	x								x	x	x	AD579K AD	(2839)	
				x	x								x	x	x	AD579T †AD	(2839)	
		2.2	755 *	x	x								x	x	x	AD579J AD	(2839)	
		5	1400 *			x	x	x					x	x	x	MN5240-10 MicroNet	(3043)	
			1600 *			x	x	x					x	x	x	DDC-5240-10 †DDC		
		6	1100			x	x	x					x	x	x	ADC-84-10 Datel	(2862)	20
						x	x	x					x	x	x	ADC-85C-10 Datel	(2862)	
			1200			x	x	x					x	x	x	DDCADC87-10 DDC		
			1200 *			x	x	x					x	x	x	DDCADC85-10 DDC		
		8	1400 *			x	x	x					x	x	x	MNADC84-10 MicroNet	(3043)	
						x	x	x					x	x	x	MNADC85-10 †MicroNet	(3043)	25
		10	1100			x	x	x					x	x	x	ADC-87-10 Datel		
			1500			x	x	x					x	x	x	ADADC84-10 AD	(2839)	
						x	x	x					x	x	x	ADC84K-10 Burr-Brown	(2848)	
			1800			x	x	x					x	x	x	ADADC85-10 AD	(2839)	
				x	x	x	x	x					x	x	x	ADC85-10 Burr-Brown	(2848)	30
		15	—	x									x			AD573K AD	(2832,2838)	
				x									x			AD573S †AD	(2832,2838)	
													x	x		ZN442E Ferranti		
													x	x		ZN442J †Ferranti		
		20	350 *	x	x								x	x	x	ZN432BJ-10 Ferranti		35
				x	x								x	x	x	ZN432CJ-10 Ferranti		
				x	x								x	x	x	ZN432E-10 Ferranti		
				x	x								x	x	x	ZN432J-10 †Ferranti		
				x	x								x	x	x	ZN432CK-10 DDC		
				x	x								x	x	x	ZN432K-10 DDC		40
		21	800			x	x	x					x	x	x	ADADC80-15 AD	(2838)	
						x	x	x					x	x	x	ADADC80Z-10 †AD	(2838)	
		30	300	x	x								x	x		AD571K AD		
		40	300	x	x									x	x	μA571K Fairchild		
				x	x									x	x	μA571T Fairchild		45
		80 *												x		AD7573K AD	(2835,2838)	
														x		AD7573T †AD	(2835,2838)	
		120	5 *	x	x									x	x	AD7570L AD		
				x	x									x	x	MP7570L MicroParr	(3046)	
		5000	20 *	x	x									x		4144 TeledyneP		50
		6000	20	x	x							x		x		ADC-EK108 Datel	(2862)	
			25	x								x		x		TSC8701 TeledyneS		
				x								x		x		TSC8704 TeledyneS		
			43	x										x		ADC-ET108M †Datel	(2862)	
			50	x										x		ADC-ET108C Datel	(2862)	55

Bin.—Binary Off.—Offset Compl.—Complementary Magn.—Magnitude CTC—Compl. 2's Compl. Int. Ref.—Internal Reference Mux. In—Multiplexed Inputs S&H—Sample and Hold Par. Out—Parallel Output Ser. Out—Serial Output

**IC MASTER**

**INTERFACE—Analog to Digital Converters (Cont'd)**

Bits Res.	Linear-ity Error $\pm$ LSB	Conver-sion Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-grating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
10	1	15	—	X									X			AD573J	AD (2832,2838)	(Cont'd)
		30	300	X	X								X	X		AD571J	AD	
				X	X								X	X		AD571S	† AD	
				X	X								X	X		MCE571J	MCE	
				X	X								X	X		MCE571S	† MCE	5
		40	300	X	X									X	X	$\mu$ A571J	Fairchild	
				X	X									X	X	$\mu$ A571S	Fairchild	
		80*												X		AD7573J	AD (2835,2838)	
														X		AD7573S	† AD (2835,2838)	
		200	—	X										X		ADC1001C	National	10
				X										X		ADC1021C	National	
10 Plus Sign	1/2	80	—	X	X		X							X	X	AD7571K	AD (2834,2838)	
				X	X		X							X	X	AD7571T	† AD (2834,2838)	
	1	80	—	X	X		X							X	X	AD7571J	AD (2834,2838)	
				X	X		X							X	X	AD7571S	† AD (2834,2838)	15
11	1	15	500	X	X								X	X		HS574J	HybridSys (2981)	
				X	X								X	X		HS574S	† HybridSys (2981)	
12	1/2	—	1500									X	X			MP2316	Analogic	
					X			X	X			X	X			MP2321	Analogic	
		2735		X	X			X		X	X	X	X			DAS5712	Intech	20
	0.5	1500		X	X								X	X		MN5247	† MicroNet (3043)	
	1/2	2500		X	X								X	X		AH8022	Analogic	
	1	1500				X	X	X					X	X	X	MN5245A	MicroNet (3043)	
	1.5	2350*			X	X							X	X	X	ADC803B	Burr-Brown (2848)	
	2	1800*		X	X			X					X	X	X	ADH-8516-12	DDC	25
		1900		X	X			X					X	X	X	ADC-817MC	Datel (2860)	
				X	X			X					X	X	X	ADC-817MM	† Datel (2860)	
		2000				X	X	X					X	X	X	MN5243	† MicroNet (3043)	
		2500				X	X	X					X	X	X	ADC00401	DDC	
		2700				X	X	X						X	X	ADC-810MC	Datel (2860)	30
						X	X	X						X	X	ADC-810MM	† Datel (2860)	
	2.2	1800*				X	X						X			HAS-1202	AD (2839)	
	3	775*		X	X								X	X		AD578L	AD (2839)	
		1900		X	X			X					X	X	X	ADC-827MC	Datel (2860)	
				X	X			X					X	X	X	ADC-827MM	† Datel (2860)	35
	3.5	1.8		X	X			X					X	X	X	ADC60-12	Burr-Brown (2848)	
	4	2700				X	X	X						X	X	ADC-811MC	Datel (2860)	
						X	X	X						X	X	ADC-811MM	† Datel (2860)	
	4.5	755*		X	X								X	X	X	AD578K	AD (2839)	
		775*		X	X								X	X	X	AD578T	† AD (2839)	40
	5	—														AD5240K	AD (2839)	
																AD5240S	AD (2839)	
		1400*				X	X	X					X	X	X	MN5240-12	MicroNet (3043)	
		1600*				X	X	X					X	X	X	DDC-5240-12	† DDC	
		2175				X	X	X					X	X	X	ADH-8586-12	† DDC	45
	6	775*		X	X								X	X	X	AD578J	AD (2839)	
				X	X								X	X	X	AD578S	† AD (2839)	
	8	1400*				X	X	X					X	X	X	MNADC84-12	MicroNet (3043)	
						X	X	X					X	X	X	MNADC85-12	MicroNet (3043)	
						X	X	X					X	X	X	MNADC87	MicroNet (3043)	50
		1500				X	X	X					X	X	X	ADC87/MIL	† Burr-Brown (2853)	
						X	X	X					X	X	X	ADC87U	Burr-Brown (2853)	
		2000				X	X	X					X	X	X	ADC-HZ12BGC	Datel (2860)	
						X	X	X					X	X	X	ADC-HZ12BMM	† Datel (2860)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

(Continued)

INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Conver-sion Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. (max.) mW	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-gra-ting	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																<b>(Cont'd)</b>		
12	1/2	9	2150			x	x				x		x	x	x	ADC-HS12BMC	Datel (2860)	
						x	x				x		x	x	x	ADC-HS12BMM $\dagger$	Datel (2860)	
		9*	2845	x	x					x	x		x	x		HDAS-16MC	Datel (2861)	
				x	x					x	x		x	x		HDAS-16MM $\dagger$	Datel (2861)	
				x	x					x	x		x	x		HDAS-8MC	Datel (2861)	5
				x	x					x	x		x	x		HDAS-8MM $\dagger$	Datel (2861)	
		10	1100			x	x	x					x	x	x	ADC-84-12	Datel (2862)	
						x	x	x					x	x	x	ADC-85-12	Datel (2862)	
						x	x	x					x	x	x	ADC-85C-12	Datel (2862)	
					x	x	x	x					x	x	x	ADC-87-12	Datel	10
		1200				x	x	x					x	x	x	HSADC85B $\dagger$	HybridSys (2983)	
						x	x	x					x	x	x	HSADC85C	HybridSys (2983)	
		1200*				x	x	x					x	x	x	DDCADC85-12	DDC	
						x	x	x					x	x	x	DDCADC87-12	DDC	
		1550				x	x	x					x	x	x	ADADC84-12	AD (2839)	15
						x	x	x					x	x	x	ADC84K-12	Burr-Brown (2848)	
		1575				x	x	x					x	x	x	ADH-8585-12 $\dagger$	DDC	
		1800				x	x	x					x	x	x	ADADC85-12	AD (2839)	
						x	x	x					x	x	x	ADC85-12	Burr-Brown (2848)	
		2000		x				x					x	x	x	HI5712-2 $\dagger$	Harris (2961)	20
				x				x					x	x	x	HI5712-5	Harris (2961)	
				x				x					x	x	x	HI5712-7	Harris (2961)	
				x				x					x	x	x	HI5712-8 $\dagger$	Harris (2961)	
				x				x					x	x	x	HI5712A-2 $\dagger$	Harris (2961)	
				x				x					x	x	x	HI5712A-7	Harris (2961)	25
				x				x					x	x	x	HI5712A-8 $\dagger$	Harris (2961)	
		13	744			x	x						x	x		ADC-5213	Datel (2860)	
						x	x						x	x		ADC-5213E	Datel (2860)	
						x	x						x	x		ADC-5213H $\dagger$	Datel (2860)	
						x	x						x	x		ADC-5214	Datel (2860)	30
						x	x						x	x		ADC-5214E	Datel (2860)	
						x	x						x	x		ADC-5214H $\dagger$	Datel (2860)	
						x	x						x	x		ADC-5215	Datel (2860)	
						x	x						x	x		ADC-5215E	Datel (2860)	
						x	x						x	x		ADC-5215H $\dagger$	Datel (2860)	35
		745		x									x	x		MN5213	MicroNet (3043)	
				x									x	x		MN5213H $\dagger$	MicroNet (3043)	
						x							x	x		MN5214	MicroNet (3043)	
						x							x	x		MN5214H $\dagger$	MicroNet (3043)	
						x							x	x		MN5215	MicroNet (3043)	40
						x							x	x		MN5215H $\dagger$	MicroNet (3043)	
		913				x	x	x					x	x		MN5610 $\dagger$	MicroNet (3043)	
		915				x	x						x	x	x	ADC-5210	Datel (2860)	
						x	x						x	x	x	ADC-5210E	Datel (2860)	
						x	x						x	x	x	ADC-5210H $\dagger$	Datel (2860)	45
						x	x						x	x	x	ADC-5211	Datel (2860)	
						x	x						x	x	x	ADC-5211E	Datel (2860)	
						x	x						x	x	x	ADC-5211H $\dagger$	Datel (2860)	
						x	x						x	x	x	ADC-5212	Datel (2860)	
						x	x						x	x	x	ADC-5212E	Datel (2860)	50
						x	x						x	x	x	ADC-5212H $\dagger$	Datel (2860)	
						x	x						x	x	x	ADC-5216	Datel (2860)	
						x	x						x	x	x	ADC-5216E	Datel (2860)	
						x	x						x	x	x	ADC-5216H $\dagger$	Datel (2860)	
						x							x	x	x	ADC582B-12 $\dagger$	HybridSys	55

Bin.—Binary  
Off.—Offset

Compl.—Complementary  
Magn.—Magnitude

CTC—Compl. 2's Compl.  
Int. Ref.—Internal Reference

Mux. In.—Multiplexed Inputs  
S&H—Sample and Hold

Par. Out.—Parallel Output  
Ser. Out.—Serial Output

**IC MASTER**

**INTERFACE-Analog to Digital Converters (Cont'd)**

Bits Res.	Linear-ity Error ± LSB	Conver-sion Time ± ½ LSB μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-grating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																<b>(Cont'd)</b>		
12	1/2	13	915															(Cont'd)
						x							x	x	x	ADC582C-12	HybridSys	
						x	x						x	x	x	DDC-5210-1	† DDC	
						x	x						x	x	x	DDC-5211-1	† DDC	
						x	x						x	x	x	DDC-5212-1	† DDC	
						x	x						x	x	x	DDC-5216-1	† DDC	5
						x	x						x	x	x	DDC5210-3	DDC	
						x	x						x	x	x	DDC5211-3	DDC	
						x	x						x	x	x	DDC5212-3	DDC	
						x	x						x	x	x	DDC5216-3	DDC	
						x	x						x	x	x	MN5210	MicroNet (3043)	10
						x	x						x	x	x	MN5210H	† MicroNet (3043)	
						x	x						x	x	x	MN5211	MicroNet (3043)	
						x	x						x	x	x	MN5211H	† MicroNet (3043)	
						x	x						x	x	x	MN5212	MicroNet (3043)	
						x	x						x	x	x	MN5212H	† MicroNet (3043)	15
						x	x						x	x	x	MN5216	MicroNet (3043)	
						x	x						x	x	x	MN5216H	† MicroNet (3043)	
						x	x						x	x	x	TP5210	† TeledyneP	
		1000		x									x	x	x	AD5210	AD (2838)	
													x	x	x	AD5211	AD (2838)	20
													x	x	x	AD5212	AD	
				x										x	x	AD5213	AD	
														x	x	AD5214	AD (2838)	
														x	x	AD5215	AD	
						x							x	x	x	AD5216	AD	25
15	500			x	x								x	x		HS574K	HybridSys (2981)	
				x	x								x	x		HS574L	HybridSys (2981)	
19	1250			x	x			x		x	x		x	x		AD/AM-822	Analogic	
20	725					x	x	x					x	x	x	ADC581B-12	† HybridSys	
						x	x	x					x	x	x	ADC581C-12	HybridSys	30
		2000				x	x	x					x	x	x	ADC-HX12BGC	Datal (2860)	
						x	x	x					x	x	x	ADC-HX12BMM	† Datal (2860)	
25	725			x	x								x		x	AD574AK	AD (2833,2838)	
				x	x								x	x		MN574A	† MicroNet (3043)	
		800				x	x	x					x	x	x	ADADC80-12	AD (2838)	35
						x	x	x					x	x	x	ADADC80Z-12	† AD (2838)	
		925		x	x			x					x	x	x	AD572A	AD (2838)	
				x	x			x					x	x	x	AD572B	AD (2838)	
				x	x			x					x	x	x	AD572S	† AD (2838)	
		950				x	x	x					x	x	x	MNADC80	MicroNet (3043)	40
		950 *				x	x	x					x	x	x	ADC80A-12	Barr-Brown (2848)	
30	780			x	x								x		x	AD574AT	† AD (2833,2838)	
				x	x								x		x	AD574K	AD (2833,2838)	
				x									x		x	AD574L	AD (2833,2838)	
				x	x								x		x	AD574S	† AD (2833,2838)	45
				x	x								x		x	AD574T	† AD (2833,2838)	
				x									x		x	AD574U	† AD (2833,2838)	
				x	x								x		x	MCD574K	MCE	
				x									x		x	MCE574L	MCE	
				x	x								x		x	MCE574S	† MCE	50
				x	x								x		x	MCE574T	† MCE	
				x									x		x	MCE574U	† MCE	
35	620			x	x								x		x	AD574ZL	AD (2833,2838)	
				x	x								x		x	AD574ZS	† AD (2833,2838)	
				x	x								x		x	AD574ZT	† AD (2833,2838)	55

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error ± LSB	Conver-sion Time ± 1/2 LSB μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Int-e-gra-ting	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
12	1/2	35	620															
				x	x								x		x	AD574ZU	† AD (2833,2838)	
				x	x								x		x	MCE574ZL	MCE	
				x	x								x		x	MCE574ZS	† MCE	
				x	x								x		x	MCE574ZT	† MCE	
				x	x								x		x	MCE574ZU	MCE	5
		50	725	x										x	x	AD504B	AD (3351)	
				x									x	x	x	AD5201B	AD	
				x									x	x	x	AD5201T	† AD	
				x									x	x	x	AD5202B	AD	
				x									x	x	x	AD5202T	† AD	10
				x										x	x	AD5204T	† AD	
				x										x	x	AD5205B	AD	
				x										x	x	AD5205T	† AD	
			745	x										x	x	MN5203	MicroNet (3043)	15
				x										x	x	MN5203H	† MicroNet (3043)	
							x							x	x	MN5204	MicroNet (3043)	
							x							x	x	MN5204H	† MicroNet (3043)	
							x							x	x	MN5205	MicroNet (3043)	
							x							x	x	MN5205H	† MicroNet (3043)	
			770	x	x			x					x	x	x	ADC10HT	† Burr-Brown (2848)	20
				x	x			x					x	x	x	ADC10HT-1	† Burr-Brown (2848)	
			915			x	x						x	x	x	DDC-5200-1	† DDC	
						x	x						x	x	x	DDC-5201-1	† DDC	
						x	x						x	x	x	DDC-5202-1	† DDC	
						x	x						x	x	x	DDC-5206-1	† DDC	25
						x	x						x	x	x	DDC5200-3	DDC	
						x	x						x	x	x	DDC5201-3	DDC	
						x	x						x	x	x	DDC5202-3	DDC	
						x	x						x	x	x	DDC5206-3	DDC	
				x									x	x	x	MN5200	MicroNet (3043)	30
				x									x	x	x	MN5200H	† MicroNet (3043)	
						x							x	x	x	MN5201	MicroNet (3043)	
						x							x	x	x	MN5201H	† MicroNet (3043)	
						x							x	x	x	MN5202	MicroNet (3043)	
						x							x	x	x	MN5202H	† MicroNet (3043)	35
						x							x	x	x	MN5206	MicroNet (3043)	
						x							x	x	x	MN5206H	† MicroNet (3043)	
			1000				x						x	x	x	AD5201	AD	
							x						x	x	x	AD5202	AD	
				x										x	x	AD5203	AD	40
							x							x	x	AD5204	AD	
							x							x	x	AD5205	AD	
							x							x	x	AD5206	AD	
			125	740									x		x	4161	† TeledyneP	45
													x		x	4162	† TeledyneP	
						x							x		x	4163	† TeledyneP	
						x	x						x		x	4164	† TeledyneP	
						x	x						x		x	4165	† TeledyneP	
			175	80			x						x	x		HS5250C	HybridSys	50
							x						x	x		HS5251C	HybridSys	
							x						x	x		HS5252C	HybridSys	
						x							x	x		HS5253C	HybridSys	
						x							x	x	x	MN5250	MicroNet	
						x							x	x	x	MN5250H	† MicroNet	
							x						x	x	x	MN5251	MicroNet	55

Bin.—Binary      Compl.—Complementary      CTC—Compl. 2's Compl.      Mux. In.—Multiplexed Inputs      Par. Out.—Parallel Output  
 Off.—Offset      Magn.—Magnitude      Int. Ref.—Internal Reference      S&H—Sample and Hold      Ser. Out.—Serial Output

INTERFACE-Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Conver-sion Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-gra-ting	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
12	1/2	175	80					x					x	x	x	MNS251H	† MicroNet	(Cont'd)
								x					x	x	x	MNS252	MicroNet	
								x					x	x	x	MNS252H	† MicroNet	
						x							x	x	x	MNS253	MicroNet	
						x							x	x	x	MNS253H	† MicroNet	5
	250	311														MNS700	MicroNet (3043)	
	300	112		x	x			x					x	x	x	ADC-HC12BMC	Datel (2860)	
				x	x			x					x	x	x	ADC-HC12BMM	† Datel (2860)	
	2400	43		x											x	ADC-ET12BM	† Datel (2862)	
	5000	15		x											x	μPD7002	NEC	10
	20000	20		x	x										x	4145	TeledyneP	
	24000	20		x	x							x		x		ADC-EK12B	Datel (2862)	
		25		x								x		x		TSC8702	TeledyneS	
				x								x		x		TSC8705	TeledyneS	
	100000	20		x					x			x	x	x	x	ICL7109	Intersil (2996)	15
				x								x	x	x	x	TSC7109BC	TeledyneS (3094)	
				x					x			x	x	x	x	TSC7109C	TeledyneS (3094)	
				x					x			x	x	x	x	TSC7109M	† TeledyneS (3094)	
3/4	1.5	2350 *			x	x							x	x	x	ADC803C	Burr-Brown (2848)	
	200	210		x											x	ADC1210	† National	20
				x											x	ADC1210C	National	
1	—	500							x			x		x		ADC-7109	Datel	
	2	2700		x	x			x					x	x	x	ADH-8516-11	DDC	
	5	1600 *				x	x	x					x	x	x	DDC-5240-12	† DDC	
	10	600		x				x					x	x		AM6112	AMD (3306)	25
	15	500		x	x								x	x		HS574T	† HybridSys (2981)	
				x	x								x	x		HS574U	† HybridSys (2981)	
		720		x	x								x	x		HI674AJ	Harris	
				x	x								x	x		HI674AS	† Harris	
	25	390		x									x			MCE574AS	† MCE	30
		720		x	x								x	x		HI574AJ	Harris (2969)	
				x	x								x	x		HI574AS	† Harris (2969)	
		725		x	x								x			AD574AJ	AD (2833,2838)	
				x	x								x			AD574AS	† AD (2833,2838)	
		750		x	x								x	x		MP574J	MicroPwr (3046)	35
				x	x								x	x		MP574S	† MicroPwr (3046)	
	30	780		x	x								x		x	AD574J	AD (2833,2838)	
				x	x								x		x	MCE574J	MCE	
	35	620		x	x								x		x	AD574ZJ	AD (2833,2838)	40
				x	x								x		x	AD574ZK	AD (2833,2838)	
				x	x								x		x	MCE574ZJ	MCE	
				x	x								x		x	MCE574ZK	MCE	
	50	725		x									x	x	x	AD5201A	AD	
				x									x	x	x	AD5201S	† AD	
				x									x	x	x	AD5202A	AD	45
				x									x	x	x	AD5202S	† AD	
				x										x	x	AD5204A	AD	
				x										x	x	AD5204S	† AD	
				x										x	x	AD5205A	AD	
				x										x	x	AD5205S	† AD	50
	175	80				x							x	x		HS5250B	† HybridSys	
							x						x	x		HS5251B	† HybridSys	
							x						x	x		HS5252B	† HybridSys	
						x							x	x		HS5253B	† HybridSys	
1/2	15	720		x	x								x	x		HI674AK	Harris	55

(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

Master Selection Guide

INTERFACE

INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Conver-sion Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-gra-ting	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line
<b>Binary Output</b>																	<b>(Cont'd)</b>	
12	1/2	15	720	x	x								x	x		HI674AL	Harris	(Cont'd)
				x	x								x	x		HI674AT	† Harris	
				x	x								x	x		HI674AU	† Harris	
		25	720	x	x								x	x		HI574AK	Harris	(2969)
				x	x								x	x		HI574AL	Harris	(2969)
				x	x								x	x		HI574AT	† Harris	(2969)
				x	x								x	x		HI574AU	† Harris	(2969)
			750	x	x								x	x		MP574K	MicroPwr	(3046)
				x	x								x	x		MP574L	MicroPwr	(3046)
				x	x								x	x		MP574T	† MicroPwr	(3046)
				x	x								x	x		MP574U	† MicroPwr	(3046)
1 1/2	2400	50		x										x		ADC-ET12BC	Datel	(2862)
2	3	2175				x	x	x					x	x	x	ADH-8586-10	† DDC	
	6	1575				x	x	x					x	x	x	ADH-8585-10	† DDC	
	200	210		x										x	x	ADC1211	† National	
				x										x	x	ADC1211C	National	
4	21	950 *				x	x	x					x	x	x	ADC80A-10	Burr-Brown	(2848)
12 Plus Sign	—	160 ms	—	x												AD7552	AD	(2838)
12 (2 device set)	1/2	15	785	x									x	x	x	HS5210B	† HybridSys	(2982)
				x									x	x	x	HS5210C	HybridSys	(2982)
				x									x	x	x	HS5211B	† HybridSys	(2982)
				x									x	x	x	HS5211C	HybridSys	(2982)
				x									x	x	x	HS5212B	† HybridSys	(2982)
				x									x	x	x	HS5212C	HybridSys	(2982)
				x										x	x	HS5213B	† HybridSys	(2982)
				x										x	x	HS5213C	HybridSys	(2982)
				x										x	x	HS5214B	† HybridSys	(2982)
				x										x	x	HS5214C	HybridSys	(2982)
				x										x	x	HS5215B	† HybridSys	(2982)
				x										x	x	HS5215C	HybridSys	(2982)
				x									x	x	x	HS5216B	† HybridSys	(2982)
				x									x	x	x	HS5216C	HybridSys	(2982)
		250000	40	x					x			x	x	x		ICL7104-12	Intersil	
			360	x					x			x	x	x		ICL8052A	Intersil	
12 (3-Digit BCD)	1/4	1200	20									x		x		ADC-EK12DC	Datel	(2862)
												x		x		ADC-EK12DM	† Datel	(2862)
	1	6000 *	50									x	x		x	AD2020	AD	
13	1/2	2	2000	x	x						x		x	x		MN5263	MicroNet	
		10	2400	x				x					x	x	x	MP2713C	Analogic	
		40000 *	64					x						x		MP7550B	MicroPwr	(3046)
			72					x				x		x	x	AD7550B	AD	(2838)
14	1/2	—	2735	x	x			x		x	x		x	x		DAS5714	Intech	
		7	2000	x				x					x	x		MP2734	Analogic	
		10	2400	x				x					x	x	x	MP2714C	Analogic	
			3000	x	x			x			x		x	x		AD/AM-724	Analogic	
				x	x			x					x	x	x	MP8014	Analogic	
		40		x										x		ICL7115	Intersil	(2997)
		50	1100	x	x			x			x		x	x		AD/AM-824	Analogic	
		60	800	x	x			x			x		x	x		AD/AM-834	Analogic	
		125 kHz									x					HAS-1409	AD	(2837)
		250	300				x						x	x	x	MN5260	MicroNet	(3043)

Bin.—Binary      Compl.—Complementary      CTC—Compl. 2's Compl.      Mux. In—Multiplexed Inputs      Par. Out—Parallel Output  
 Off.—Offset      Magn.—Magnitude      Int. Ref.—Internal Reference      S&H—Sample and Hold      Ser. Out—Serial Output



INTERFACE—Analog to Digital Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Conver-sion Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Output	Off. Bin. Output	Compl. Bin. Output	Compl. Off. Bin. Output	CTC or 2's Compl. Output	Sign. Magn. Output	Mux. In.	S&H	Inte-grating	Int. Ref.	Par. Out	Ser. Out	Device	Source	Line	
<b>Binary Output</b>																		<b>(Cont'd)</b>	
<b>14 (2 device set)</b>																			
1/2	250000	40		x					x			x	x	x		ICL7104-14	Intersil		
		360		x					x			x	x	x		ICL8052A	Intersil		
<b>14 (3 1/2-Digit BCD)</b>																			
1/2	10000	20												x		4146	TeledyneP		
<b>15</b>																			
1/2	5	1750						x					x	x		MP2735-2	Analogic		
	8	1750						x			x		x	x		MP2735-1	Analogic	5	
	15	3000		x	x			x					x	x	x	MP8015	Analogic		
	50	1100		x	x			x			x		x	x		AD/AM-825	Analogic		
	60	800		x	x			x			x		x	x		AD/AM-835	Analogic		
2	400000	20		x					x						x	TSC800AC	TeledyneS (3098)		
				x					x						x	TSC800AM	† TeledyneS (3098)	10	
4	400000	20		x					x						x	TSC800BC	TeledyneS (3098)		
				x					x						x	TSC800BM	† TeledyneS (3098)		
<b>16</b>																			
1/2	—	2735		x	x			x		x	x		x	x		DAS5716	Intech		
	8	3000		x	x			x					x	x	x	A8016-9	Intech		
	10	3000		x	x			x					x	x	x	A8016-10	Intech	15	
	16	3000		x	x			x					x	x	x	A8016-16	Intech		
	20	3000		x	x			x					x	x	x	A8016	Intech		
	2	3500		x	x			x			x		x	x		AD/AM-826	Analogic		
	2.8	1000 *				x	x	x					x	x	x	ADC82A	Barr-Brown (2848)		
	32	3000		x	x			x					x	x	x	MP8016	Analogic	20	
	100	1200		x	x			x						x		HS9516-4	HybridSys (2982)		
	170	2500		x	x			x					x	x	x	ADC731K	Barr-Brown (2848)		
				x	x			x					x	x	x	ADC73K	Barr-Brown (2848)		
	3000	750									x	x			x	MP2344	Analogic		
1	50	—		x		x	x	x							x	ADADC71K	AD (2836,2838)	25	
				x		x	x	x							x	ABADC72K	AD (2836,2838)		
	100	1200		x	x			x						x		HS9516-5	HybridSys (2982)		
	170	2500		x	x			x					x	x	x	ADC731J	Barr-Brown (2848)		
				x	x			x					x	x	x	ADC73J	Barr-Brown (2848)		
1 *	17	1550				x	x	x					x	x	x	PCM75K	Barr-Brown (2848)	30	
2	15	1550 *				x	x	x					x	x		ADC76K	Barr-Brown (2848)		
	40	1000				x	x	x					x	x	x	MN5290	† MicroNet (3043)		
						x	x	x					x	x	x	MN5291	† MicroNet (3043)		
	50	—		x		x	x	x						x	x	ADADC71J	AD (2836,2838)		
				x		x	x	x						x	x	ADADC72J	AD (2836,2838)	35	
		1550 *				x	x	x					x	x	x	ADC71K	Barr-Brown (2848)		
						x	x	x					x	x		ADC72B	Barr-Brown (2848)		
	60	200				x	x	x					x	x	x	MN5284	MicroNet (3043)		
	100	1200		x	x			x						x		HS9516-6	HybridSys (2982)		
2 *	17	1550				x	x	x					x	x	x	PCM75J	Barr-Brown (2848)	40	
4	15	1550 *				x	x	x					x	x		ADC76J	Barr-Brown (2848)		
	50	1440				x	x	x					x	x	x	MN5282	MicroNet (3043)		
		1550 *				x	x	x					x	x		ADC72A	Barr-Brown (2848)		
						x	x	x					x	x		ADC72J	Barr-Brown (2848)		
						x	x	x					x	x		ADC72K	Barr-Brown (2848)	45	
		1800 *				x	x	x					x	x	x	ADC71J	Barr-Brown (2848)		
	100	1440				x	x	x					x	x	x	MN5280	MicroNet (3043)		
<b>16 (2 device set)</b>																			
1/2	250000	40		x					x			x	x	x		ICL7104-16	Intersil		
		360		x					x			x	x	x		ICL8052A	Intersil		
<b>17</b>																			
1/2	400	1550							x			x	x	x		MP8037	Analogic	50	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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Master Selection Guide

INTERFACE—Analog to Digital Converters (Cont'd)

Digits	Device	Source	Line
<b>Decimal Output</b>			
3 1/2 Digits, Integrating	<b>ICL7106</b>	<b>Intersil (2995)</b>	5
	<b>ICL7107</b>	<b>Intersil (2995)</b>	
	ICL7116	Intersil	
	ICL7117	Intersil	
	ICL7126	Intersil	
	<b>ICL7136</b>	<b>Intersil (3005)</b>	
	ICL7137	Intersil	
	<b>MP7138</b>	<b>MicroPwr (3046)</b>	
	<b>MP7138A</b>	<b>MicroPwr (3046)</b>	
	MC14433	Motorola	
	ADC3511	National	
	ADD3501	National	
	TSC1433A	TeledyneS	
	TSC14433	TeledyneS	
	TSC14433B	TeledyneS	
	TSC7106	TeledyneS	
	TSC7106A	TeledyneS	
	TSC7107	TeledyneS	
	TSC7107A	TeledyneS	
	TSC7116	TeledyneS	
TSC7116A	TeledyneS		
TSC7117	TeledyneS		
TSC7126	TeledyneS		
TSC7126A	TeledyneS		
3 Digits, Dual Slope, Building Block, for Microprocessor Systems, e.g. TMS1000	TL505C	TI	25
3 Digits, Dual Slope, 2 Device Sets	<b>CA3161</b>	<b>RCA (3600)</b>	25
	<b>CA3162</b>	<b>RCA (3600)</b>	
3 1/2 Digits, Drives LCD DVM Display	ZN450	Ferranti	30
	ZN451	Ferranti	
3 1/2 Digits, Integrating, 2 Device Sets	ICL7101	Intersil	35
	ICL7103	Intersil	
	ICL8052	Intersil	
	ICL8053	Intersil	
	ICL8068	Intersil	
	<b>LD110</b>	<b>Siliconix (3083)</b>	
<b>LD111A</b>	<b>Siliconix (3083)</b>		
3 1/2 Digits, Dual Slope	TSC8751	TeledyneS	40
3 1/2—4 1/2 Digits, Ramp type, 2 Device Sets	MC1405	Motorola	40
	MC14435	Motorola	
	MC14435E	† Motorola	
	MC1505	† Motorola	
3 1/2	TSC8750	TeledyneS	45
3 3/4 Digits, Integrating	ZNA216E	Ferranti	45
	ZNA216J	† Ferranti	
	ADC3711	National	
	ADD3701	National	
4 1/2 Digits, Integrating, for microprocessor or UART systems	SI7135	<b>Siliconix (3083)</b>	50
4 1/2 Digits, Successive Integration	<b>ICL7129</b>	<b>Intersil (2999)</b>	50
4 1/2 Digits, Dual Slope	<b>ICL7135</b>	<b>Intersil (3003)</b>	50
	<b>TSC7135</b>	<b>TeledyneS (3095)</b>	
4 1/2 Digits, Dual Slope, 2 Device Sets	ICL7103A	Intersil	55
	ICL8053A	Intersil	
	ICL8068A	Intersil	
	<b>LD120</b>	<b>Siliconix (3083)</b>	
	<b>LD121A</b>	<b>Siliconix (3083)</b>	
<b>LD122</b>	<b>Siliconix (3083)</b>		

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INTERFACE-Digital to Analog Converters

Master Selection Guide

Bits Res.	Linear-ity Error ± LSB	Settling Time ± ½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																		
4	1/4	0.3	85		x					x	X		x			ZN434E	Ferranti	
	1/2	0.004				x						x			x	DAC-0405	Datel	
			1040			x						x	x		x	<b>MN0405</b>	<b>MicroMet</b>	<b>(3044)</b>
4 (triple color video DAC)																		
	1/2	0.005 *	1250	x						x					x	VDAC444	Intech	
4 (triple color video RAM/DAC)																		
	1/2	0.005 *	3000	x						x					x	RGBDAC4T	Intech	5
4 (video digital converter)																		
	1/2	0.004 *	1040				x					x			x	<b>HDG-0405</b>	<b>AD</b>	<b>(2830)</b>
						x						x			x	0405H	Intech	
6	1/8	3	250			x	x			x			x			<b>DAC-01A</b>	<b>† PMI</b>	<b>(3567)</b>
	1/4	3	250			x	x			x			x			DAC-01	MicroPwr	
						x	x			x			x			DAC-01B	† MicroPwr	10
						x	x			x			x			DAC-01C	MicroPwr	
						x	x			x			x			DAC-01D	MicroPwr	
						x	x			x			x			DAC-01F	† MicroPwr	
						x	x			x			x			DAC-01H	MicroPwr	
						x	x	x		x			x			<b>MP5520A</b>	<b>† MicroPwr</b>	<b>(3046)</b>
						x	x	x		x			x			<b>MP5520B</b>	<b>† MicroPwr</b>	<b>(3046)</b>
						x	x	x		x			x			<b>MP5520C</b>	<b>MicroPwr</b>	<b>(3046)</b>
						x	x	x		x			x			<b>MP5520F</b>	<b>† MicroPwr</b>	<b>(3046)</b>
						x	x	x		x			x			<b>MP5520H</b>	<b>MicroPwr</b>	<b>(3046)</b>
						x	x									μPC603	NEC	20
						x	x			x			x			<b>DAC-01</b>	<b>† PMI</b>	<b>(3567)</b>
						x	x			x			x			<b>DAC-01B</b>	<b>† PMI</b>	<b>(3567)</b>
						x	x			x			x			<b>DAC-01C</b>	<b>PMI</b>	<b>(3567)</b>
						x	x			x			x			<b>DAC-01F</b>	<b>† PMI</b>	<b>(3567)</b>
						x	x			x			x			<b>DAC-01H</b>	<b>PMI</b>	<b>(3567)</b>
	1/2	0.006				x						x			x	DAC-0605	Datel	
			1352			x						x	x		x	<b>MN0605</b>	<b>MicroMet</b>	<b>(3044)</b>
		0.3	240			x	x							x		<b>MC1406</b>	<b>Motorola</b>	<b>(3069)</b>
						x	x							x		<b>MC1506</b>	<b>† Motorola</b>	<b>(3069)</b>
	1 *	45 *		x						x	x		x			ZN426E-6	Ferranti	30
	3	200				x	x	x		x			x			<b>MP5520D</b>	<b>MicroPwr</b>	<b>(3046)</b>
		250				x	x			x			x			<b>DAC-01D</b>	<b>PMI</b>	<b>(3567)</b>
6 (A/D,D/A,with counter)																		
	1/2	2.0	175	x						x	x		x			ZN425E-6	Ferranti	
6 (video digital converter)																		
	1/2	0.006 *	1350			x						x			x	<b>HDG-0605</b>	<b>AD</b>	<b>(2830)</b>
			1500			x						x			x	0605H	Intech	35
6/12-Binary Serial																		
	1/2	—	5							x	x			x	x	μA9706C	Fairchild	
7	1/2	1 *	45 *	x						x	x		x			ZN426E-7	Ferranti	
				x						x	x		x			ZN429E-7	Ferranti	
7 (A/D,D/A,with counter)																		
	1/2	2.0	175	x						x	x		x			ZN425E-7	Ferranti	
8	—	0.005	—	x								x	x	x		<b>AD9768</b>	<b>† AD</b>	<b>(2823,2830)</b>
	1/8	0.15	1.5 *	x	x	x	x				x			x		<b>AD7523L</b>	<b>AD</b>	<b>(2825)</b>
				x	x	x	x				x			x		AD7523L	Intersil	
				x	x	x	x				x			x		AD7523U	† Intersil	
				x	x	x	x				x			x		<b>MP7523L</b>	<b>MicroPwr</b>	<b>(3046)</b>
			20	x	x	x	x				x			x	x	<b>MP7524C</b>	<b>MicroPwr</b>	<b>(3046)</b>
				x	x	x	x			x	x			x	x	<b>AD7524C</b>	<b>AD</b>	<b>(2825)</b>
				x	x	x	x			x	x			x	x	<b>AD7524L</b>	<b>AD</b>	<b>(2825)</b>
				x	x	x	x			x	x			x	x	<b>AD7524U</b>	<b>† AD</b>	<b>(2825)</b>
				x	x	x	x				x			x	x	<b>MP7524L</b>	<b>MicroPwr</b>	<b>(3046)</b>
				x	x	x	x				x			x	x	<b>MP7524U</b>	<b>† MicroPwr</b>	<b>(3046)</b>

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INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Settling Time $\mu$ S	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
8	1/8	0.15 *	670	x	x	x	x				x				x	AD7523J	Intersil	(Cont'd)
				x	x	x	x				x				x	AD7523S	† Intersil	
	1	30		x						x	x				x	DAC0830	National	
	1 *	—		x						x	x		x			ZN438	Ferranti	
	1/4	0.01	850	x								x				HDS-0810E	AD (2830)	5
				x								x				HDS-0810EM	† AD (2830)	
				x								x				MC10318-9	Motorola (3069)	
		0.045	700	x							x	x				HI5618A-2	† Harris (2929)	
				x							x	x				HI5618A-5	Harris (2929)	
		0.065	610	x	x			x		x	x					4084-02	† TeledyneP	10
				x	x			x		x	x					4084-03	† TeledyneP	
		0.085	33	x	x	x	x			x	x	x				DAC-080QF	AMD	
		0.135	174	x	x	x	x			x	x	x				DAC-08A	† AMD	
				x	x	x	x			x	x	x				ADDAC08A	† AD (2825)	15
					x	x	x			x	x	x				ADDAC08H	AD (2825)	
				x	x	x	x			x	x	x				DAC-08A	† Motorola (3069)	
				x	x	x	x			x	x	x				DAC-08H	Motorola (3069)	
				x	x	x	x			x						DAC0802	† National	
				x	x	x	x			x						DAC0802C	National	
				x	x	x	x			x	x	x				DAC-08A	† PMI	20
				x	x	x	x			x	x	x				DAC-08H	PMI	
				x	x	x	x			x	x	x				DAC-08A	† Raytheon (3591)	
				x	x	x	x			x	x	x				DAC-08H	Raytheon (3591)	
				x	x	x	x			x	x	x				DAC-08A	† Signetics	
				x	x	x	x			x	x	x				DAC-08H	Signetics	25
	0.15	1.5 *		x	x	x	x				x					AD7523K	AD (2825)	
				x	x	x	x				x					AD7523K	Intersil	
				x	x	x	x				x					AD7523T	† Intersil	
				x	x	x	x				x					MP7523K	MicroPwr (3046)	
	20			x	x	x	x				x				x	MP7524B	MicroPwr (3046)	30
	0.15 *	20 *		x	x					x	x				x	AD7524B	AD (2825)	
				x	x	x	x			x	x				x	AD7524K	AD (2825)	
				x	x	x	x			x	x				x	AD7524T	† AD (2825)	
				x	x	x	x				x				x	MP7524K	MicroPwr (3046)	
				x	x	x	x				x				x	MP7524T	† MicroPwr (3046)	35
	0.16 *	123(5V)		x	x	x	x	x		x	x				x	AM6080AC	AMD	
				x	x	x	x	x		x	x				x	AM6080AM	† AMD	
				x	x	x	x	x	x	x	x				x	AM6081AC	AMD	
				x	x	x	x	x	x	x	x				x	AM6081AM	† AMD	
	0.2 *	255 *		x						x	x	x	x		x	NE5119	Signetics (3611)	40
				x						x	x	x	x		x	SE5119	† Signetics (3611)	
		1300 *				x	x				x	x			x	HDH-0802	AD (2830)	
						x	x				x	x			x	HDH-0802M	† AD (2830)	
	0.25	170		x	x					x					x	DAC-888A	† PNI (3567)	45
				x	x					x					x	DAC-888E	PNI (3567)	
	0.25 *	265		x	x					x	x				x	AD1408-9	AD (2825)	
				x	x					x	x				x	AD1508-9	† AD (2825)	
	1	30		x						x	x					DAC0831	National	
	1.5	375		x						x	x		x		x	AD558K	AD (2817,2825)	
	2 *	255		x						x	x	x	x	x	x	NE5019	Signetics (3607)	50
				x						x	x	x	x	x	x	SE5019	† Signetics (3607)	
	20 *	735 *		x	x						x	x				HDS-0820	AD (2830)	
				x	x						x	x				HDS-0820M	† AD (2830)	

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ± LSB	Settling Time ± ½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
8																		
	3/8	1.5	375	x						x	x				x	<b>AD558T</b>	† AD (2817,2825)	
	1/2									x					x	<b>AD7548K</b>	<b>AD</b> (2822,2827)	
										x					x	<b>AD7548T</b>	† AD (2822,2827)	
	0.005	—										x			x	SP9768	Plessey	
	0.008					x						x			x	DAC-0805	Datel	5
		800		x								x				<b>TDC1018</b>	<b>TRW-LSI</b> (3702)	
		1664				x						x	x		x	<b>MN0805</b>	<b>MicroNet</b> (3044)	
	0.01	700		x		x				x					x	<b>TDC1016-8</b>	<b>TRW-LSI</b> (3704)	
	0.01*	675		x								x				<b>MC10318</b>	<b>Motorola</b> (3069)	
	0.025	630		x	x					x			x			<b>DAC-HF8BMM</b>	† Datel (2860)	10
				x	x					x			x			DAC-HFBMC	Datel	
	0.045	700		x						x	x					<b>HI5618B-2</b>	† Harris (2929)	
				x						x	x					<b>HI5618B-5</b>	Harris (2929)	
	0.065	610		x	x			x		x	x					4084	TeledyneP	
				x	x			x		x	x					4084-01	TeledyneP	15
	0.12	240				x	x			x			x			<b>DAC90S</b>	† Burr-Brown (2848)	
		255		x						x	x	x	x	x	x	NE5118	Signetics	
				x						x	x	x	x	x	x	SE5118	† Signetics	
	0.135	174		x	x	x	x			x	x	x			x	DAC-08	† AMD	
				x	x	x	x			x	x	x			x	<b>ADDAC08</b>	† AD (2825)	20
				x	x	x				x	x	x			x	DAC08	† Fairchild	
				x	x	x	x			x	x	x			x	<b>DAC-08</b>	† Motorola (3069)	
				x	x	x				x					x	DAC0800	† National	
				x	x	x	x			x	x	x			x	DAC-08	† PMI	
				x	x	x	x			x	x	x			x	<b>DAC-08</b>	† Raytheon (3591)	25
				x	x	x	x			x	x	x			x	DAC-08	† Signetics	
	0.14*	130*		x	x					x			x		x	<b>MC6890A</b>	† Motorola (3068)	
	0.15	1.5*		x	x	x	x				x				x	<b>AD7523J</b>	<b>AD</b> (2825)	
				x	x	x	x				x				x	<b>AD7523S</b>	† AD (2825)	
				x	x	x	x				x				x	<b>MP7523J</b>	<b>MicroPwr</b> (3046)	30
				x	x	x	x				x				x	<b>MP7524A</b>	<b>MicroPwr</b> (3046)	
		20		x	x					x					x	<b>DAC-08BC</b>	<b>Datel</b> (2863)	
		136		x	x					x	x	x			x	<b>DAC-08BM</b>	† Datel (2863)	
				x	x					x	x	x			x	DAC-08E	AMD	
		174		x	x	x	x			x	x	x			x	<b>ADDAC08E</b>	<b>AD</b> (2825)	35
				x	x	x	x			x	x	x			x	DAC08E	Fairchild	
				x	x	x	x			x	x	x			x	<b>DAC-08E</b>	<b>Motorola</b> (3069)	
				x	x	x				x					x	DAC0800C	National	
				x	x	x				x					x	μPC624	NEC	
				x	x	x	x			x	x	x			x	DAC-08E	PMI	40
				x	x	x	x			x	x	x			x	<b>DAC-08E</b>	<b>Raytheon</b> (3591)	
				x	x	x	x			x	x	x			x	DAC-08E	Signetics	
	0.15*	20*		x	x					x	x				x	<b>AD7524A</b>	<b>AD</b> (2825)	
				x	x	x	x				x				x	<b>AD7524J</b>	<b>AD</b> (2825)	
				x	x	x	x				x				x	<b>AD7524S</b>	† AD (2825)	45
				x	x	x	x				x				x	<b>MP7524J</b>	<b>MicroPwr</b> (3046)	
				x	x	x	x				x				x	<b>MP7524S</b>	† <b>MicroPwr</b> (3046)	
		305		x						x	x				x	DAC0808	† National	
				x						x	x				x	DAC0808C	National	
	0.16*	123		x	x	x	x	x		x	x				x	<b>AM6080C</b>	AMD	50
				x	x	x	x	x		x	x				x	<b>AM6080M</b>	† AMD	
				x	x	x	x	x	x	x	x				x	<b>AM6081C</b>	AMD	
				x	x	x	x	x	x	x	x				x	<b>AM6081M</b>	† AMD	
	0.2	240				x	x			x			x			<b>DAC90B</b>	<b>Burr-Brown</b> (2848)	
		670		x	x					x	x				x	<b>DAC-7523</b>	<b>Datel</b> (2863)	55

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Settling Time $\mu$ s	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
8	1/2	0.20 *	130 *	x	x					x			x		x	MC6890	Motorola (3068)	(Cont'd)
		0.22	20 *	x	x					x	x			x	x	AD7528K	AD (2819,2825)	
				x	x					x	x			x	x	AD7528L	AD (2819,2825)	
				x	x					x	x			x	x	AD7528T	† AD (2819,2825)	
				x	x					x	x			x	x	AD7528U	† AD (2819,2825)	5
		0.25	170	x	x					x				x	x	DAC-888B	† PMI (3567)	
				x	x					x				x	x	DAC-888F	PMI (3567)	
		0.25 *	265	x						x	x			x		SSS1408A-8	AMD	
				x						x	x			x		SSS1508A-8	† AMD	
				x	x					x	x			x		AD1408-8	AD (2825)	10
				x	x					x	x			x		AD1508-8	† AD (2825)	
				x	x					x	x			x		DAC-1408A-8	PMI	
				x	x					x	x			x		DAC-1508A-8	† PMI	
		0.25/1.0																
			900				x	x	x	x				x	x	DAC82S	† Barr-Browne (2848)	
		0.25/2.0																
			900				x	x	x	x				x	x	DAC82K	Barr-Browne (2848)	15
		0.3 *	305	x						x	x			x		AM1408-8	AMD	
				x						x	x			x		AM1508-8	† AMD	
				x	x					x	x			x		DAC-IC88C	Datal (2863)	
				x	x					x	x			x		DAC-IC88M	† Datal (2863)	
				x						x	x			x		DAC1408	† Fairchild	20
				x						x	x			x		DAC1408A	Fairchild	
				x										x		MC1408-8	Motorola (3069)	
				x										x		MC1508-8	† Motorola (3069)	
				x						x				x		MC1408-8	National	
				x						x				x		MC1508-8	† National	25
				x						x				x		MC1408-8	Signetics	
				x						x				x		MC1508-8	† Signetics	
		0.6 *	500	x			x			x				x		XR9201	Exar	
		0.8	100 *	x	x					x	x			x		ZN428E-8	Ferranti	
				x	x					x	x			x		ZN428J-8	† Ferranti	30
		1	30	x						x	x			x		DAC0832	National	
			45 *	x						x	x			x		ZN426E-8	Ferranti	
				x						x	x			x		ZN426J-8	† Ferranti	
				x						x	x					ZN429E-8	Ferranti	
				x						x	x					ZN429J-8	† Ferranti	35
			750	x						x				x		MN3008	MicroNet (3044)	
				x						x				x		MN3008H	† MicroNet (3044)	
					x					x				x		MN3009	MicroNet (3044)	
					x					x				x		MN3009H	† MicroNet (3044)	
		1.5	75 *	x						x	x			x		AD558J	AD (2817,2825)	40
			300	x	x					x	x			x		DAC331B-8	† HybridSys	
				x	x					x	x			x		DAC331C-8	HybridSys	
		2 *	255	x	x					x				x		DAC-UP88C	Datal (2863)	
				x	x					x				x		DAC-UP88M	† Datal (2863)	
			255 *	x						x	x	x	x	x	x	NE5018	Signetics	45
				x						x	x	x	x	x	x	SE5018	† Signetics (526)	
		2.5	570	x	x	x				x				x		MN3014	MicroNet (3044)	
				x	x	x				x				x		MN3014H	† MicroNet (3044)	
		3	680	x	x									x		HS3020B	† HybridSys	
				x	x									x		HS3020C	HybridSys	50
			830	x	x	x				x				x		MN3020	MicroNet (3044)	
				x	x	x				x				x		MN3020H	† MicroNet (3044)	
		4	1080	x	x			x		x	x			x		DAC336B-8	† HybridSys	
				x	x			x		x	x			x		DAC336C-8	HybridSys	

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ±LSB	Settling Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																		<b>(Cont'd)</b>	
8	1/2																<b>(Cont'd)</b>		
	20	285		x						x	x		x			DAC336B-1	† HybridSys		
				x						x	x		x			DAC337B-0	† HybridSys		
				x						x	x		x			DAC337B-2	† HybridSys		
				x						x	x		x			DAC337C-0	HybridSys		
				x						x	x		x			DAC337C-1	HybridSys	5	
				x						x	x		x			DAC337C-2	HybridSys		
	23 *	585				x				x	x		x			MN3000	MicroNet (3044)		
						x				x	x		x			MN3000H	† MicroNet (3044)		
					x					x	x		x			MN3001	MicroNet (3044)		
					x					x	x		x			MN3001H	† MicroNet (3044)	10	
				x						x	x		x			MN3002	MicroNet (3044)		
				x						x	x		x			MN3002H	† MicroNet (3044)		
					x					x	x		x			MN3006	MicroNet (3044)		
					x					x	x		x			MN3006H	† MicroNet (3044)		
	30	570		x	x	x				x			x			MN3013	MicroNet (3044)	15	
				x	x	x				x			x			MN3013H	† MicroNet (3044)		
	40	300			x					x	x		x			DAC337B-6	† HybridSys		
					x					x	x		x			DAC337C-6	HybridSys		
3/4	1.5	375		x						x	x		x		x	AD558S	† AD (2817,2825)		
1										x				x		AD7548J	AD (2822,2827)	20	
										x				x		AD7548S	† AD (2822,2827)		
	0.01	—		x	x					x					x	TM1840	Telmos		
	0.01 *	675		x								x		x		MC10318C-7	Motorola (3069)		
	0.07	305		x						x	x			x		MC1408-7	Signetics		
	0.15	174		x	x	x	x			x	x	x		x		DAC-08C	AMD	25	
				x	x	x	x			x	x	x		x		ADDAC08C	AD (2825)		
				x	x	x	x			x	x	x		x		DAC08C	Fairchild		
				x	x	x	x			x	x	x		x		DAC-08C	Motorola (3069)		
				x	x	x				x				x		DAC0801C	National		
				x	x	x	x			x	x	x		x		DAC-08C	PMI	30	
				x	x	x	x			x	x	x		x		DAC-08C	Raytheon (3501)		
				x	x	x	x			x	x	x		x		DAC-08C	Signetics		
	0.22	20 *		x	x					x	x			x	x	AD7528J	AD (2819,2825)		
				x	x					x	x			x	x	AD7528S	† AD (2819,2825)		
	0.25 *	265		x	x					x	x			x		SSS1408A-7	AMD	35	
				x	x					x	x			x		AD1408-7	AD (2825)		
				x	x					x	x			x		DAC1408B	Fairchild		
				x	x					x	x			x		DAC-1408A-7	PMI		
	0.3 *	305		x						x	x			x		AM1408-7	AMD		
				x										x		MC1408-7	Motorola (3069)	40	
				x						x				x		DAC0807C	National		
				x						x				x		MC1408-7	National		
1 1/4	0.3 *	305		x							x			x		MC3408	Motorola (3069)		
2	0.01 *	675		x								x		x		MC10318C-6	Motorola (3069)		
	0.25 *	265		x	x					x	x			x		DAC1408C	Fairchild	45	
				x	x					x	x			x		DAC-1408A-6	PMI		
	0.3 *	305		x						x	x			x		AM1408-6	AMD		
				x										x		MC1408-6	Motorola (3069)		
				x						x				x		DAC0806C	National		
				x						x				x		MC1408-6	National	50	
8 (A/D, D/A, with counter)	1/2	2.0	175	x						x	x		x			ZN425E-8	Ferranti		
	1	2.0	175	x						x	x		x			ZN425J-8	† Ferranti		
8 (D/A, with counter/clock)	1/2	0.8	—							x	x		x			ZN435E	Ferranti		
										x	x		x			ZN435J	† Ferranti		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linearity Error $\pm$ LSB	Settling Time $\mu$ s	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																	<b>(Cont'd)</b>		
<b>8 (triple color video RAM/DAC)</b>																			
1/2	0.008 *	8000		x								x			x	RGBDAC8E	Intech		
		10000		x						x					x	RGBDAC8T	Intech		
<b>8 (video digital converter)</b>																			
1/4	0.010	1976		x								x				HDD-0810	AD	5	
				x								x				HDD-0810M	† AD		
		2028		x								x				HDD-0810C	AD		
				x								x				HDD-0810CM	† AD		
1/2	0.008	1660				x						x				HDE-0805	AD (2830)		
	0.008 *	1800				x						x			x	0805H	Intech		
	0.07	1000		x								x			x	AH8308E	Analogic	10	
				x						x						AH8308T	Analogic		
		1600		x								x			x	MP8308ECL	Analogic		
				x								x			x	MP8318ECL	Analogic		
		2200		x						x					x	MP8308	Analogic		
				x						x					x	MP8318	Analogic		
<b>8 (Dual), Data Latches, Addressable</b>																			
1	—	1350		x									x		x	MP10	Barr-Brown (2849)	15	
				x									x		x	MP11	Barr-Brown (2849)		
<b>8 (7-Bits plus Sign) Companding</b>																			
—	0.5 *	192																	
																x	AM6072C	AMD	
																x	AM6072M	† AMD	
1/2 step	0.5	192														x	AM6070AC	AMD	
																x	AM6070AM	† AMD	20
										x	x	x			x	DAC-88E	PMI (3567)		
		207								x	x	x			x	DAC-88E	PMI (3567)		
		500								x	x	x			x	DAC-89E	PMI (3567)		
1 step	0.5	192														x	AM6070C	AMD	
																x	AM6070M	† AMD	25
	0.5 *	192								x	x	x			x	DAC-86C	PMI (3567)		
		207								x	x	x			x	DAC-88C	PMI (3567)		
		500								x	x	x			x	DAC-89C	PMI (3567)		
<b>8-2 Digit BCD</b>																			
1/2	0.15	194								x	x	x			x	DAC-20C	PMI (3567)		
9	1/2	0.01	700	x		x				x						TDC1016-9	TRW-LSI (3704)	30	
<b>9 (8-Bits plus Sign)</b>																			
1/4	0.75	500		x					x	x	x		x	x		DAC-208A	† PMI (3567)		
				x					x	x	x		x	x		DAC-208E	PMI (3567)		
1/2	0.75	500		x					x	x	x		x	x		DAC-208B	† PMI (3567)		
				x					x	x	x		x	x		DAC-208F	PMI (3567)		
10	1/8	0.5	20	x	x	x	x		x	x				x		MP7533R	MicroPwr (3046)	35	
				x	x	x	x		x	x				x		MP7533Y	† MicroPwr (3046)		
1/4 *	0.6 *	780		x					x	x				x		HI5610-2	† Harris (2926)		
				x					x	x				x		HI5610-4	Harris (2926)		
				x					x	x				x		HI5610-5	Harris (2926)		
				x					x	x				x		HI5610-8	† Harris (2926)	40	
1/4	0.025 *	735		x	x				x	x						HDS-1025	AD (2830)		
				x	x				x	x						HDS-1025M	† AD (2830)		
	0.25 *	275		x	x				x	x			x			AD561K	AD (2825)		
				x	x				x	x			x			AD561T	† AD (2825)		
	0.3 *	1300				x	x		x	x			x			HDH-1003	AD (2830)	45	
	0.5	20		x	x	x	x		x	x				x		MP7533Q	MicroPwr (3046)		
				x	x	x	x		x	x				x		MP7533W	† MicroPwr (3046)		
1/2	0.01	450		x		x				x					x	TDC1016-10	TRW-LSI (3704)		
	0.025	705		x	x					x						DAC-HF10BMC	Datal (2860)		
				x	x					x						DAC-HF10BMM	† Datal (2860)	50	
	0.15	450		x	x	x	x			x	x	x			x	DAC-10B	† PMI (3567)		

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying



**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ±LSB	Settling Time ± 1/2 LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																		<b>(Cont'd)</b>	
10	1/2	0.15	450	x	x	x	x			x	x	x			x	DAC-10C	† PMI (3567)	(Cont'd)	
				x	x	x	x			x	x	x			x	DAC-10F	PMI (3567)		
				x	x	x	x			x	x	x			x	DAC-10G	PMI (3567)		
	0.25	380									x				x	MC3410	Motorola (3069)	5	
											x				x	MC3510	† Motorola (3069)		
											x				x	MC3410	Signetics		
		390		x	x					x	x					DAC-IC10B	Datel (2863)		
				x	x					x	x					DAC-IC10BM	† Datel (2863)		
	0.25 *	275		x	x					x	x		x			AD561J	AD (2825)	10	
				x	x					x	x		x			AD561S	AD (2825)		
	0.3	—		x	x						x				x	AD7527C	AD (2825)		
				x	x						x				x	AD7527G	AD (2825)		
				x	x						x				x	AD7527GL	AD (2825)		
				x	x						x				x	AD7527GU	† AD (2825)		
				x	x						x				x	AD7527L	AD (2825)	15	
				x	x						x				x	AD7527U	† AD (2825)		
	0.375	300				x	x			x			x			DAC-100A	† PMI (3567)		
						x	x			x			x			DAC-100A	PMI		
	0.5	30		x	x					x	x				x	DAC-HA10BC-1	Datel (2860)	20	
				x	x					x	x				x	DAC-HA10BM-1	† Datel (2860)		
	0.5 *	0.025		x	x					x	x				x	DAC-HA10BC	Datel (2860)		
				x	x					x	x				x	DAC-HA10BM	† Datel (2860)		
	20 *			x	x	x	x			x	x				x	AD7520L	AD (2825)	25	
				x	x	x	x			x	x				x	AD7520U	† AD (2825)		
				x	x	x	x			x	x				x	AD7530L	AD (2825)		
				x	x	x	x			x	x				x	AD7520L	Intersil		
				x	x	x	x			x	x				x	AD7520U	† Intersil		
				x	x	x	x			x	x				x	AD7530L	Intersil		
				x	x	x	x			x	x				x	MP7520L	MicroPwr (3046)	30	
				x	x	x	x			x	x				x	MP7520U	† MicroPwr (3046)		
				x	x	x	x			x	x				x	MP7530	MicroPwr (3046)		
	20 *			x						x	x				x	DAC1000	† National	35	
				x						x	x				x	DAC1000C	National		
				x	x	x	X			x	x				x	DAC1006	† National		
				x	x	x	X			x	x				x	DAC1006C	National		
	40			x	x	x	X			x	x				x	AD7522L	AD (2825)		
				x	x	x	X			x	x				x	AD7522U	† AD (2825)		
				x	x					x	x				x	MP7522L	MicroPwr (3046)		
				x	x	x	x			x	x				x	MP7522U	† MicroPwr (3046)		
	0.6	30		x	x	x	x			x	x				x	AD7533C	AD (2825)	40	
				x	x	x	x			x	x				x	AD7533L	AD (2825)		
				x	x	x	x			x	x				x	AD7533U	† AD (2825)		
	0.8 *	30 *		x	x	x	x			x	x				x	AD7533L	Intersil	45	
				x	x	x	x			x	x				x	AD7533U	† Intersil		
				x	x	x	x			x	x				x	MP7533L	MicroPwr (3046)		
				x	x	x	x			x	x				x	MP7533U	† MicroPwr (3046)		
	1.5	30		x	x					x	x				x	DAC331B-10	† HybridSys		
				x	x					x	x				x	DAC331C-10	HybridSys		
	4 *	255		x						x	x	x	x	x	x	NE5020	Signetics (3608)	50	
	10	715				x	x			x			x			MN3040	MicroNet (3044)		
						x	x			x			x			MN3040H	† MicroNet (3044)		
	15	165		x				x		x	x				x	DAC348B-10	† HybridSys		
				x				x		x	x				x	DAC348C-10	HybridSys		
	20	150				x	x			x	x				x	DAC347LPB-10B	† HybridSys		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Settling Time ±½ LSB µs	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
10	1/2	20	150			X				X	X		X			DAC347LPB-10U † HybridSys		(Cont'd)
						X	X			X	X		X			DAC347LPC-10B HybridSys		
						X				X	X		X			DAC347LPC-10U HybridSys		
		285			X					X	X		X			DAC337B-4 † HybridSys		
				X						X	X		X			DAC337B-5 † HybridSys		5
					X					X	X		X			DAC337C-4 HybridSys		
				X						X	X		X			DAC337C-5 HybridSys		
23 *		585				X				X			X			MN3003 MicroNet (3044)		
						X				X			X			MN3003H † MicroNet (3044)		
					X					X			X			MN3004 † MicroNet (3044)		10
					X					X			X			MN3004H † MicroNet (3044)		
				X						X			X			MN3005 MicroNet (3044)		
				X						X			X			MN3005H † MicroNet (3044)		
					X					X			X			MN3007 MicroNet (3044)		
					X					X			X			MN3007H † MicroNet (3044)		15
30		—			X					X				X		M/D2005-1 HyComp		
40		300				X				X	X		X			DAC337B-3 † HybridSys		
					X					X	X		X			DAC337B-7 † HybridSys		
						X				X	X		X			DAC337C-3 HybridSys		
					X					X	X		X			DAC337C-7 HybridSys		20
500 *		24		X						X	X			X		AD7520L National		
				X						X	X			X		AD7520U † National		
				X						X	X			X		DAC1020 † National		
				X						X	X			X		DAC1020C National		
1/2		0.01	990	X								X				HDS-1015E AD (2830)		25
				X								X				HDS-1015EM † AD (2830)		
		0.012		X								X	X	X		SP9770 Plessey (3071)		
		0.135	276	X		X				X	X			X		DAC-10B Raytheon (3587)		
				X		X				X	X			X		DAC-10C Raytheon (3587)		
				X		X				X	X			X		DAC-10F Raytheon (3587)		30
		0.25	380	X						X	X			X		MC9510 Signetics		
		0.25 *	300	X						X	X		X	X		NE5410 Signetics (3612)		
				X						X	X		X	X		SE5410 Signetics (3612)		
		0.3 *	1300			X	X			X	X		X			HDH-1003M † AD (2830)		
		0.375	250			X	X	X		X			X			ADDAC100K AD (2825)		35
						X	X	X		X			X			ADDAC100L AD (2825)		
						X	X	X		X			X			ADDAC100T † AD (2825)		
		0.5	20	X	X	X	X			X	X			X		MP7520R MicroPwr (3046)		
				X	X	X	X			X	X			X		MP7520Y † MicroPwr (3046)		
		0.8	670	X	X					X	X			X		DAC-7533 Datal (2863)		40
1		0.01	700	X		X				X					X	TDC1016-9 TRW-LSI		
		0.15	276	X		X				X	X			X		DAC-10G Raytheon (3587)		
		0.25	380	X							X			X		MC3410C Meterata (3069)		
				X							X			X		MC3410C Signetics		
			390	X	X					X	X			X		DAC-1C10BC Datal (2863)		45
		0.3	—	X	X						X			X		AD7527B AD (2825)		
				X	X						X			X		AD7527K AD (2825)		
				X	X						X			X		AD7527T † AD (2825)		
			300			X	X			X			X			DAC-100B † PMI (3567)		
		0.375	250			X	X	X		X			X			ADDAC100J AD (2825)		50
						X	X	X		X			X			ADDAC100S † AD (2825)		
		0.5	20	X	X	X	X			X	X			X		MP7520Q MicroPwr (3046)		
				X	X	X	X			X	X			X		MP7520W † MicroPwr (3046)		
		0.5 *	20 *	X	X	X	X			X	X			X		AD7520K AD (2825)		

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ±LSB	Settling Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																		<b>(Cont'd)</b>	
10	1	0.5*	20*	x	x	x	x			x	x				x	AD7520T †AD	(2825)	(Cont'd)	
				x	x	x	x			x	x				x	AD7530K AD	(2825)		
				x	x	x	x			x	x				x	AD7520K Intersil			
				x	x	x	x			x	x				x	AD7520T † Intersil			
				x	x	x	x			x	x				x	AD7530K Intersil		5	
				x	x	x	x			x	x				x	MP7520K MicroPwr	(3046)		
				x	x	x	x			x	x				x	MP7520T † MicroPwr	(3046)		
				x	x	x	x			x	x				x	MP7530K MicroPwr	(3046)		
				x	x	x	x			x	x				x	DAC1001 † National			
				x	x	x	x			x	x				x	DAC1001C National		10	
				x	x	x	x			x	x				x	DAC1007 † National			
				x	x	x	x			x	x				x	DAC1007C National			
			40	x	x	x	x			x	x				x	AD7522K AD	(2825)		
				x	x	x	x			x	x				x	AD7522T †AD	(2825)		
			50	x	x	x	x			x	x				x	MP7522K MicroPwr	(3046)	15	
				x	x	x	x			x	x				x	MP7522T † MicroPwr	(3046)		
			0.6	30	x	x	x	x		x	x				x	AD7533B AD	(2825)		
				x	x	x	x			x	x				x	AD7533K AD	(2825)		
				x	x	x	x			x	x				x	AD7533T †AD	(2825)		
			0.8*	30*	x	x	x	x		x	x				x	AD7533K Intersil		20	
				x	x	x	x			x	x				x	AD7533T † Intersil			
				x	x	x	x			x	x				x	MP7533K MicroPwr	(3046)		
			1.5*	300		x		x		x	x				x	DAC-06E PMI	(3567)		
				350	x					x	x				x	DAC-03AD PMI	(3567)		
				x						x	x				x	DAC-03BD PMI	(3567)	25	
			5	435	x	x									x	DAC-UP10B Datal	(2863)		
			500*	24	x					x	x				x	AD7520K National			
				x						x	x				x	AD7520T National			
				x						x	x				x	DAC1021 † National			
				x						x	x				x	DAC1021C National		30	
2	0.01	700		x		x				x					x	TDC1016-8 TRW-LSI			
	0.225	300				x	x			x					x	DAC-100C † PMI	(3567)		
						x	x			x					x	DAC-100C PMI			
	0.5*	20*		x	x	x	x			x	x				x	AD7520J AD	(2825)	35	
				x	x	x	x			x	x				x	AD7520S †AD	(2825)		
				x	x	x	x			x	x				x	AD7530J AD	(2825)		
				x	x	x	x			x	x				x	AD7520J Intersil			
				x	x	x	x			x	x				x	AD7520S † Intersil			
				x	x	x	x			x	x				x	AD7530J Intersil			
				x	x	x	x			x	x				x	MP7520J MicroPwr	(3046)	40	
				x	x	x	x			x	x				x	MP7520S MicroPwr	(3046)		
				x	x	x	x			x	x				x	MP7530J MicroPwr	(3046)		
				x	x	x	x			x	x				x	DAC1002 † National			
				x	x	x	x			x	x				x	DAC1002C National			
							x			x	x					DAC1008 † National		45	
										x	x				x	DAC1008C National			
			40	x	x	x	x			x	x				x	AD7522J AD	(2825)		
				x	x	x	x			x	x				x	AD7522S †AD	(2825)		
			50	x	x	x	x			x	x				x	MP7522J MicroPwr	(3046)		
				x	x	x	x			x	x				x	MP7522S † MicroPwr	(3046)	50	
			0.6	30	x	x	x	x		x	x				x	AD7533A AD	(2825)		
				x	x	x	x			x	x				x	AD7533J AD	(2825)		
				x	x	x	x			x	x				x	AD7533S †AD	(2825)		
			0.8*	30*	x	x	x	x		x	x				x	AD7533J Intersil		(Continued)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value  
**Bold face indicates additional data is provided on the page noted.**

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Settling Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
10	2	0.8 *	30 *															
				x	x	x				x	x				x	AD7533S	† Intersil	
				x	x	x	x			x	x				x	MP7533S	† MicroPwr (3046)	
				x	x	x	x			x	x				x	MP7533T	† MicroPwr (3046)	
				x	x	x	x			x	x				x	MP7533T	MicroPwr	
		1.5 *	300					x		x	x				x	DAC-06F	PMI (3567)	5
			350	x						x	x				x	DAC-03CD	PMI (3567)	
		500 *	24	x						x	x				x	AD7520J	National	
				x						x	x				x	AD7520S	† National	
				x						x	x				x	DAC1022	† National	
				x						x	x				x	DAC1022C	National	10
	3	0.225	300			x	x			x					x	DAC-100D	† PMI (3567)	
						x	x			x					x	DAC-100D	PMI	
		0.5 *	50	x	x	x	x			x	x				x	MP7520H	MicroPwr (3046)	
		1.5 *	350		x			x		x	x				x	DAC-06B	PMI (3567)	
	4	0.2	—		x	x				x	x				x	DAC-10G	PMI	15
		1.5 *	300		x			x		x	x				x	DAC-06B	PMI (3567)	
			350	x						x	x				x	DAC-03DD	PMI (3567)	
	5	1.5 *	350					x		x	x				x	DAC-06C	† PMI (3567)	
	8	0.5 *	50	x	x	x	x			x	x				x	MP7520G	MicroPwr (3046)	
<b>10 Companding 1 Step</b>																		
		0.5	260						x							DAC-78F	PMI	20
<b>10 Plus Sign</b>																		
	1/2	1.5 *	350	x					x	x	x				x	DAC-05E	PMI (3567)	
			500	x					x	x	x				x	DAC-210A	† PMI (3567)	
				x					x	x	x				x	DAC-210B	† PMI (3567)	
				x					x	x	x				x	DAC-210E	PMI (3567)	
	6		300	x												μPC610	NEC	25
	1	1.5 *	300	x					x	x	x				x	DAC-210G	PMI (3567)	
			350	x					x	x	x				x	DAC-02AC	PMI (3567)	
				x					x	x	x				x	DAC-02BC	PMI (3567)	
			500	x					x	x	x				x	DAC-210F	PMI (3567)	
	2	1.5 *	300	x					x	x	x				x	DAC-02CC	PMI (3567)	30
				x					x	x	x				x	DAC-05F	PMI (3567)	
			350	x					x	x	x				x	DAC-05A	† PMI (3567)	
	3	1.5 *	350	x					x	x	x				x	DAC-05B	† PMI (3567)	
				x					x	x	x				x	DAC-05C	PMI (3567)	
	4	1.5 *	300	x					x	x	x				x	DAC-05G	PMI (3567)	35
		2.5 *	350	x					x	x	x				x	DAC-02DD	PMI (3567)	
<b>10 Video Digital Converter</b>																		
	1/2	0.015	2340	x											x	HDD-1015	AD	
				x											x	HDD-1015C	AD	
				x											x	HDD-1015M	† AD	
	1/2	75 MHz		x									x	x	x	SP9770	Plessey	40
<b>11 Plus Sign Dynamic Range (7-Bit plus sign format)</b>																		
	1/2 step	0.5	207							x	x	x			x	DAC-88E	PMI	
										x	x	x			x	DAC-89E	PMI	
		0.5 *	207							x	x	x			x	DAC-86E	PMI	
	1 step	0.5	207							x	x	x			x	DAC-86C	PMI	
										x	x	x			x	DAC-88C	PMI	45
										x	x	x			x	DAC-89C	PMI	
	12	1/8	0.15	375	x	x		x		x	x					4089-03	TeledyneP	
					x	x		x		x	x					4089-05	TeledyneP	
		1/4	0.01							x	x				x	AD667K	AD (2818,2827)	
			0.15	375	x	x		x		x	x					4089-02	TeledyneP	50
					x	x		x		x	x					4089-04	TeledyneP	

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ± 1/2 LSB	Settling Time ± 1/2 LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
12	1/4																(Cont'd)	
	0.25	345		x	x					x	x		x			<b>AD565AK</b>	<b>AD (2826,2828)</b>	
				x	x					x	x		x			<b>AD565AT</b>	<b>†AD (2826,2828)</b>	
	0.3	270		x						x	x			x		DAC1266A	National	
				x						x	x			x		DAC1266AC	National	
		345		x						x	x		x	x		DAC1265A	National	
				x						x	x		x	x		DAC1265AC	National	
	800					x	x	x		x	x					TDDAC80I	TeledyneP	
	0.4	—		x	x					x	x					<b>DAC-562M</b>	<b>†Datal (2863)</b>	
	210			x	x					x	x		x			<b>MC3512</b>	<b>Motorola (3067)</b>	
	300			x	x					x	x		x	x		<b>AD566AK</b>	<b>AD (2827,2829)</b>	10
				x	x					x	x		x	x		<b>AD566AT</b>	<b>†AD (2827,2829)</b>	
				x	x					x	x		x			<b>AD566K</b>	<b>AD (2827,2829)</b>	
				x	x					x	x		x			<b>AD566T</b>	<b>†AD (2827,2829)</b>	
				x	x					x	x		x			<b>MCE566AK</b>	<b>MCE</b>	
				x	x					x	x		x	x		<b>MCE566AT</b>	<b>†MCE</b>	15
				x	x					x	x		x	x		<b>MCE566K</b>	<b>MCE</b>	
				x	x					x	x		x			<b>MCE566T</b>	<b>†MCE</b>	
	345			x	x					x	x		x			<b>AD565K</b>	<b>AD (2826,2828)</b>	
				x	x					x	x		x			<b>AD565T</b>	<b>†AD (2826,2828)</b>	
				x	x					x	x		x			<b>μA565K</b>	<b>Fairchild</b>	20
				x	x					x	x		x			<b>μA565T</b>	<b>†Fairchild</b>	
				x	x					x	x		x			<b>MCE565K</b>	<b>MCE</b>	
				x	x					x	x		x			<b>MCE565T</b>	<b>†MCE</b>	
	780			x						x	x			x		<b>HI562A-2</b>	<b>†Harris (2920)</b>	
				x						x	x			x		<b>HI562A-B</b>	<b>Harris (2920)</b>	25
	0.5	375		x	x					x	x		x			<b>HI565AK</b>	<b>Harris (2923)</b>	
				x	x					x	x		x			<b>HI565AT</b>	<b>†Harris (2923)</b>	
	495			x	x					x	x		x	x		<b>AD567K</b>	<b>AD (2827)</b>	
	1	64 *		x	x					x	x			x		<b>HI7541K</b>	<b>Harris (2953)</b>	
				x	x					x	x			x		<b>HI7541T</b>	<b>†Harris (2953)</b>	30
	1.5	465		x	x					x	x			x		<b>AD562S/BIN</b>	<b>AD (2827,2829)</b>	
				x	x					x	x			x		<b>AD562S/BIN</b>	<b>MicroPwr</b>	
	475			x	x					x	x		x			<b>AD563K/BIN</b>	<b>AD (2826)</b>	
				x	x					x	x		x			<b>AD563S/BIN</b>	<b>†AD (2826)</b>	
				x	x					x	x		x			<b>AD563T/BIN</b>	<b>†AD (2826)</b>	35
	800					x	x	x		x	x					TDDAC80V	TeledyneP	
1/2	—	35			x			x	x	x					x	<b>AD7545C</b>	<b>AD (2821,2827,2827)</b>	
					x			x	x	x					x	<b>AD75456C</b>	<b>AD (2821,2827,2827)</b>	
					x			x	x	x					x	<b>AD75456L</b>	<b>AD (2821,2827,2827)</b>	
					x			x	x	x					x	<b>AD75456U</b>	<b>AD (2821,2827,2827)</b>	40
					x			x	x	x					x	<b>AD7545L</b>	<b>AD (2821,2827,2827)</b>	
					x			x	x	x					x	<b>AD7545U</b>	<b>AD (2821,2827,2827)</b>	
	1000			x	x	x	x			x	x			x	x	<b>MP7622B</b>	<b>MicroPwr (3046)</b>	
	1000			x	x	x	x			x	x			x	x	<b>MP7622T</b>	<b>†MicroPwr (3046)</b>	
	—	1200		x	x	x	x			x	x			x	x	<b>MP7622K</b>	<b>MicroPwr (3046)</b>	45
	0.01									x	x			x		<b>AD667J</b>	<b>AD (2818,2827)</b>	
										x	x			x		<b>AD667S</b>	<b>†AD (2818,2827)</b>	
	0.035	—										x	x			<b>DAC63</b>	<b>Burr-Brown (2848)</b>	
	855			x	x					x	x					<b>HDS-1250</b>	<b>AD (2830)</b>	
				x	x					x	x					<b>HDS-1250M</b>	<b>†AD (2830)</b>	50

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Settling Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
12	1/2	0.045	1160			x	x					x	x			DAC63S	Burr-Brown (2848)	
		0.05	—			x	x			x		x	x			ADH-030II-12	DDC	
			1485			x	x					x	x			HC4000	HyComp	
						x	x					x	x			HC4000-883	† HyComp	
		0.050	780	x	x					x			x			DAC-HF12BMC	Datel (2861)	5
				x	x					x			x			DAC-HF12BMM	† Datel (2861)	
			1200	x	x					x				x		DAC391B-12	† HybridSys	
				x	x					x				x		DAC391C-12	HybridSys	
			1350			x	x					x	x			DAC397B-12	† HybridSys	
						x	x					x	x			DAC397C-12	HybridSys	10
		0.06/1	675			x	x			x			x			DAC-8528-12	† DDC	
		0.060	395	x	x					x			x			4065	† TeledyneP	
		0.1	600	x						x				x		AM6082C	AMD	
		0.10	775			x	x			x			x			DDC-DAC87-CBI-I	† DDC	
			850			x	x	x		x			x			DDC-DAC85LD-CBI-I	DDC	15
		0.18	—	x	x					x	x			x		AD7544B	AD (2827)	
				x	x					x	x			x		AD7544BG	AD (2827)	
				x	x					x	x			x		AD7544GK	AD (2827)	
				x	x					x	x			x		AD7544GT	† AD (2827)	
				x	x					x	x			x		AD7544K	AD (2827)	20
				x	x					x	x			x		AD7544T	† AD (2827)	
		0.20	430 *	x	x			x		x	x					DAC10HT	† Burr-Brown (2848)	
				x	x			x		x	x			x		DAC10HT-1	† Burr-Brown (2848)	
		0.25	345	x	x					x	x		x			AD565AJ	AD (2826,2828)	
				x	x					x	x		x			AD565AS	† AD (2826,2828)	25
				x	x					x	x		x			MCE565AJ	MCE	
				x	x					x	x		x			MCE565AS	† MCE	
			900			x	x			x			x			4080	† TeledyneP	
						x	x			x			x			4080-83	† TeledyneP	
						x	x			x			x			4081	TeledyneP	30
						x	x			x			x			4081-83	† TeledyneP	
						x	x			x			x			4082	TeledyneP	
						x	x			x			x			4082-83	† TeledyneP	
		0.3	270	x						x	x			x		DAC1266L	National	
				x						x	x			x		DAC1266LC	National	35
			345	x						x	x		x	x		DAC1265L	National	
				x						x	x		x	x		DAC1265LC	National	
			770 *			x				x	x		x			DAC-85C-CBI-I	Datel	
						x				x	x		x			DAC-87C-CBI-I	† Datel (2861,2861)	
			1000			x				x			x			HI5680I-5	Harris (2932)	40
						x				x	x		x			HI5685I-4	Harris (2938)	
						x	x			x			x			HI5687I-2	† Harris	
			1200			x	x	x			x		x			DAC80Z-CBI-I	Burr-Brown (2848)	
		0.3 *	800 *			x	x	x		x			x			DAC850-CBI-I	Burr-Brown (2848)	
						x	x	x		x			x			DAC851-CBI-I	† Burr-Brown	45
		0.3/1.5 *	800 *			x	x	x		x			x			DAC800-CBI-I	Burr-Brown (2848)	
						x	x	x		x			x			DAC800-CBI-V	Burr-Brown (2848)	
						x	x	x		x			x			DAC851-CBI-I	† Burr-Brown (2848)	
			800			x	x	x		x			x			DAC80/CBI	Burr-Brown (2848)	
						x	x	x		x			x			DAC85C-CBI-I	Burr-Brown (2848)	50
			850			x	x	x		x			x			ADDAC87/CBI	† AD (2826)	
						x	x	x		x			x			DAC85-CBI	MicroNet (3044)	
			925			x	x	x		x			x			ADDAC80/CBI	AD (2825)	

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ± LSB	Settling Time ± 1/2 LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																		<b>(Cont'd)</b>	
12	1/2	0.3/3 *	925															(Cont'd)	
																		<b>ADDAC85/CBI † AD (2825)</b>	
																		<b>ADDAC85C/CBI AD (2825)</b>	
	0.4	—		x	x					x	x							<b>DAC-562C Datal (2863)</b>	
						x	x	x										HIDAC801 Harris	
	210			x	x						x	x						<b>MC3412 Motorola (3067)</b>	
	300			x	x						x	x						<b>AD566AJ AD (2827,2829)</b>	
				x	x						x	x						<b>AD566AS † AD (2827,2829)</b>	
				x	x						x	x						<b>AD566J AD (2827,2829)</b>	
				x	x						x	x						<b>AD566S † AD (2827,2829)</b>	
				x	x						x	x						MCE565J MCE	
				x	x						x	x						MCE565S † MCE	
				x	x						x	x						MCE566AJ MCE	
				x	x						x	x						MCE566AS † MCE	
				x	x						x	x						MCE566J MCE	
				x	x						x	x						MCE566S † MCE	
	345			x	x						x	x						<b>AD565J AD (2826,2828)</b>	
				x	x						x	x						<b>AD565S † AD (2826,2828)</b>	
				x	x						x	x						μA565J Fairchild	
				x	x						x	x						μA565S † Fairchild	
	400			x	x						x	x						DAC862S † Burr-Brown	
				x	x	x	x				x	x						μPC648 NEC	
	780			x							x	x						<b>HI562A-4 Harris (2920)</b>	
				x							x	x						<b>HI562A-5 Harris (2920)</b>	
	0.5	0.05		x	x						x	x						<b>DAC-HA12BC Datal (2860)</b>	
				x	x						x	x						<b>DAC-HA12BM † Datal (2860)</b>	
	20			x	x						x	x						<b>DAC-HA12BC-1 Datal (2860)</b>	
				x	x						x	x						<b>DAC-HA12BM-1 † Datal (2860)</b>	
	312 *			x	x	x	x	x			x	x	x					<b>AM6012C AMD (3301)</b>	
				x	x	x	x	x			x	x	x					<b>AM6012M † AMD (3301)</b>	
				x	x	x	x	x			x	x	x					MCE6012C MCE	
				x	x	x	x	x			x	x	x					MCE6012M † MCE	
	375			x	x						x	x						<b>HI565AJ Harris (2923)</b>	
				x	x						x	x						<b>HI565AS † Harris (2923)</b>	
	495			x	x						x	x						<b>AD567J AD (2827)</b>	
				x	x						x	x						<b>AD567S † AD (2827)</b>	
	1000			x							x	x						<b>AD562 Motorola (3069)</b>	
				x							x	x						<b>AD563 Motorola (3069)</b>	
	0.5 *	1300				x	x				x	x						<b>HDH-1205 AD (2830)</b>	
						x	x				x	x						<b>HDH-1205M † AD (2830)</b>	
	0.6	1410		x	x						x							DAC392B-12 HybridSys	
				x	x						x							DAC392C-12 HybridSys	
	1/2	20		x	x	x	x				x	x						<b>MP7541T † MicroPwr (3046)</b>	
	1	20		x	x	x	x				x	x						<b>MP7541B MicroPwr (3046)</b>	
				x	x	x	x				x	x						<b>MP7541K MicroPwr (3046)</b>	
	20 *			x	x	x	x				x	x						<b>AD7541B AD (2825,2827,3351)</b>	
				x	x	x	x				x	x						<b>AD7541K AD (2825,2827,3351)</b>	
				x	x	x	x				x	x						<b>AD7541T † AD (2825,2827,3351)</b>	
				x	x	x	x				x	x						AD7541B Intersil	
				x	x	x	x				x	x						AD7541K Intersil	
				x	x	x	x				x	x						AD7541T Intersil	
				x	x	x	x				x	x						<b>MP7621B MicroPwr (3046)</b>	
				x	x	x	x				x	x						<b>MP7621K MicroPwr (3046)</b>	
				x	x	x	x				x	x						<b>MP7621T † MicroPwr (3046)</b>	

† Military Temperature Range (-55° to 125°C)

\* Typical Value  
**Bold face indicates additional data is provided on the page noted.**

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Settling Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
12	1/2	1	20 *															
				x	x	x	x			x	x				x	7541T	† TeledyneS	
				x	x	x	x			x	x				x	TSC7541B	TeledyneS	
				x	x	x	x			x	x				x	TSC7541K	TeledyneS	
				x	x	x	x			x	x				x	TSC8641B	† TeledyneS	
				x	x	x	x			x	x				x	TSC8641C	TeledyneS	5
		30		x	x					x	x				x	HS7541B-2	† HybridSys	
				x	x					x	x				x	HS7541C-2	HybridSys	
				x						x	x				x	DAC1208	National	
				x						x	x				x	DAC1218	National	
				x						x	x				x	DAC1230	National	10
		30 *		x	x					x	x				x	HS7541-4	HybridSys	
		64 *		x	x					x	x				x	HI7541J	Harris (2953)	
				x	x					x	x				x	HI7541S	† Harris (2953)	
		450		x	x					x	x				x	DAC-7541	Datel (2863)	
				x	x	x	x			x	x				x	MP7623B	MicroPwr (3046)	15
				x	x	x	x			x	x				x	MP7623K	MicroPwr (3046)	
				x	x	x	x			x	x				x	MP7623T	† MicroPwr (3046)	
				x	x					x	x					TDP7541B	TeledyneP	
				x	x					x	x					TDP7541K	TeledyneP	
				x	x					x	x					TDP7541T	† TeledyneP	20
		1 *	750		x	x				x	x				x	HSDAC80CBI-I	HybridSys	
		1/1.5 *	—			x	x			x					x	HSDAC87BI/V	† HybridSys (2986)	
						x	x			x					x	HSDAC87CI/V	HybridSys (2986)	
		1.5	65	x	x					x	x	x			x	MP562H	MicroPwr (3046)	
				x	x					x	x	x			x	MP562J	MicroPwr (3046)	25
				x	x					x	x	x			x	MP562K	MicroPwr (3046)	
			300			x	x			x	x				x	DAC335B-12	† HybridSys	
						x	x			x	x				x	DAC335C-12	HybridSys	
		1.5 *	—	x						x	x				x	AD563J	AD (2826)	
			465	x	x					x	x				x	AD562K/BIN	AD (2827,2829)	30
			475	x	x					x	x				x	AD563J/BIN	AD (2826)	
			750		x	x				x	x				x	HSDAC80CBI-V	HybridSys	
			800 *			x	x	x		x					x	DAC850-CBI-V	Burr-Brown (2848)	
						x	x	x		x					x	DAC851-CBI-V	Burr-Brown (2848)	
		2	40	x	x					x	x				x	AD7542B	AD (2827)	35
				x	x					x	x				x	AD7542K	AD (2827)	
				x	x					x	x				x	AD7542T	† AD (2827)	
				x						x					x	AD7543B	AD (2827)	
				x						x					x	AD7543K	AD (2827)	
			375	x	x					x					x	4058	† TeledyneP	40
			375 *	x	x	x				x	x				x	HS3120B-2	† HybridSys	
				x	x	x				x	x				x	HS3120C-2	HybridSys	
			1900			x	x			x						HDD-1206J	AD (2830)	
						x	x			x						HDD-1206S	† AD (2830)	
		2 *	—			x	x			x						DAC345I-12	HybridSys	45
		2.5	—	x	x					x	x				x	DAC338B-12-2	HybridSys (2983)	
		3	30	x	x					x	x				x	DAC331B-12	† HybridSys	
				x	x					x	x				x	DAC331C-12	HybridSys	
			770 *			x				x	x				x	DAC-85C-CBI-V	Datel	
						x				x	x				x	DAC-87C-CBI-V	† Datel (2861,2861)	50
			775			x	x			x					x	DDC-DAC87-CBI-I	† DDC	
			850			x	x	x		x					x	DDC-DAC85LD-CBI-V	DDC	
						x	x			x					x	DAC80	MicroNet (3044)	
						x	x	x		x					x	MNDAC87	† MicroNet (3044)	

Bin.—Binary  
Off.—Offset  
Magn.—Magnitude  
Compl.—Complementary  
Int Ref.—Internal Reference  
CTC—Compl. 2's Compl.  
Mult.—Multiplying

INTERFACE  
Master Selection Guide



**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ± 1/2 LSB	Setting Time ± 1/2 LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
12	1/2	3	900	x	x					x			x		x	DAC-HK12BGC	Datel (2861)	(Cont'd)
								x		x			x		x	DAC-HK12BGC-2	Datel (2861)	
				x	x					x			x		x	DAC-HK12BMM	† Datel (2861)	
								x		x			x		x	HAC-HK12BMM-2	† Datel	
				x	x					x			x		x	DAC-HK12BGC	Intech	5
				x	x					x			x		x	DAC-HK12BMM	† Intech	
			1000			x				x			x			HI5680V-5	Harris (2932)	
						x				x	x		x			HI5685V-4	Harris (2938)	
						x	x			x			x			HI5687V-2	Harris	
			1050			x	x			x			x			DAC-HZ12BGC	Datel (2861)	10
						x	x			x			x			HAC-HZ12BMM	† Datel	
3.5	465			x	x					x	x			x		AD562A/BCD	AD (2827,2829)	
				x	x					x	x			x		AD562A/BIN	AD (2827,2829)	
4	300			x	x					x	x		x		x	DAC336B-12	† HybridSys	
				x	x					x	x		x		x	DAC336C-12	HybridSys	15
	975					x	x			x			x		x	DACHK	† MicroNet (3044)	
5	450 *			x	x					x	x		x			HS9338-2	HybridSys (2983)	
	1000				x	x							x			HS3860B	† HybridSys	
					x	x							x			HS3860C	HybridSys	
	1200					x	x	x			x		x			DAC80-CBI-V	Barr-Brown (2848)	20
						x	x	x			x		x			DAC80Z-CBI-V	Barr-Brown (2848)	
						x	x	x		x			x			DAC85C-CBI-V	Barr-Brown (2848)	
						x	x	x			x		x			DAC85L-CBI-V	Barr-Brown (2848)	
						x	x	x			x		x			DAC85Z-CBI-V	Barr-Brown (2848)	
5 *	—					x	x			x						DAC345V-12	HybridSys	25
	525 *					x	x			x	x		x			MN3850	† MicroNet (3044)	
						x	x			x	x		x			MN3850H	MicroNet (3044)	
	675 *					x	x			x			x		x	MN3860	MicroNet (3044)	
						x	x			x			x		x	MN3860H	† MicroNet (3044)	
7	—			x						x	x		x	x	x	ICL7146	Intersil (3010)	30
	1000					x	x			x			x		x	AD3860K	AD (2827)	
						x	x			x			x		x	AD3860S	† AD (2827)	
	1025					x	x	x		x			x			DAC87/CBI	† Barr-Brown (2853)	
8	—			x						x	x					AD290T	† AD	
				x						x	x					AD390K	AD (2816,2827)	35
	375					x	x			x			x			MN3348	MicroNet (3044)	
						x	x			x			x			MN3348H	† MicroNet (3044)	
10	375					x	x			x			x			MN3349	MicroNet (3044)	
						x	x			x			x			MN3349H	† MicroNet (3044)	
	720			x	x			x		x			x			MP1812A	Analogic	40
	750					x	x	x		x			x		x	DAC-SL-12	† DDC	
	760					x	x			x			x		x	DAC88	† MicroNet (3044)	
						x	x		x				x		x	MN3660	† MicroNet (3044)	
	1200					x	x			x			x			DAC71	MicroNet (3044)	
15	165			x				x		x	x			x		DAC348B-12	† HybridSys	45
				x				x		x	x			x		DAC348C-12	HybridSys	
	300			x	x					x	x		x			DAC349B-12	† HybridSys	
				x	x					x	x		x			DAC349C-12	HybridSys	
				x	x					x	x		x			DAC9349-12	HybridSys	
20	150					x	x			x	x		x			DAC347LPB-12U	† HybridSys	50

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Setting Time ±1/2 LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																		(Cont'd)
12	1/2	20	150				X	X		X	X		X			DAC347LPC-12B	HybridSys	(Cont'd)
							X	X		X	X		X			DAC347LPC-12U	HybridSys	
							X	X		X	X		X			HAC347LPB-12B	HybridSys	
																† HybridSys		
		355					X			X				X		MN3412	MicroNet (3044)	
25	—							X		X	X		X		X	DAC02701	DDC	5
		175 *					X			X	X		X			DAC9356	HybridSys (2986)	
30	—			X			X			X					X	M/DA2000	HyComp	
35	90						X			X			X			MN371	MicroNet (3044)	
							X			X			X			MN371H	† MicroNet (3044)	
		150						X		X	X		X			AD370J	AD (2826)	10
								X		X	X		X			AD370K	AD (2826)	
								X		X	X		X			AD370S	† AD (2826)	
							X			X	X		X			AD371J	AD (2826)	
35 *	—			X	X					X	X		X			DDC1250-12-1	DDC	15
				X	X					X	X		X			DDC1250-12-3	DDC	
50	265						X			X	X		X			DAC356B-12	† HybridSys	
							X			X	X		X			DAC356C-12	HybridSys	
		908					X			X	X		X			DAC356LPB-12	† HybridSys	
							X			X	X		X			DAC356LPC-12	HybridSys	
		1653					X	X				X	X			HDS-1240E	AD (2830)	20
							X	X				X	X			HDS-1240EM	† AD (2830)	
70	90						X			X	X		X			MN370	MicroNet (3044)	
							X			X	X		X			MN370H	† MicroNet (3044)	
3/4	8	—		X						X	X					AD390J	AD (2816,2827)	25
				X						X	X					AD390S	† AD (2816,2827)	
1	—	35			X			X	X	X					X	AD7545B	AD (2821,2827,2827)	
					X			X	X	X					X	AD7545K	† AD (2821,2827,2827)	
					X			X	X	X					X	AD7545T	AD (2821,2827,2827)	
0.06/1	675						X	X		X			X			DAC-8528-11	† DDC	
0.18	—			X	X					X	X			X		AD7544A	AD (2827)	30
				X	X					X	X			X		AD7544J	AD (2827)	
				X	X					X	X			X		AD7544S	AD (2827)	
1	20			X	X	X	X			X	X			X		MP7541A	MicroPwr (3046)	
				X	X	X	X			X	X			X		MP7541J	MicroPwr (3046)	
				X	X	X	X			X	X			X		MP7541S	† MicroPwr (3046)	35
		20 *		X	X	X	X			X	X			X		AD7541A	AD (2825,2827,3351)	
				X	X	X	X			X	X			X		AD7541J	AD (2825,2827,3351)	
				X	X	X	X			X	X			X		AD7541A	Intersil	
				X	X	X	X			X	X			X		AD7541J	Intersil	
				X	X	X	X			X	X			X		MP7621A	MicroPwr (3046)	40
				X	X	X	X			X	X			X		MP7621J	MicroPwr (3046)	
				X	X	X	X			X	X			X		MP7621S	† MicroPwr (3046)	
				X	X	X				X	X			X		7541S	† TeledyneS	
				X	X	X				X	X			X		TSC7441S	† TeledyneS	
				X	X	X				X	X			X		TSC7541A	TeledyneS	45
				X	X	X				X	X			X		TSC7541J	TeledyneS	
				X	X	X				X	X			X		TSC8640B	† TeledyneS	
				X	X	X				X	X			X		TSC8640C	TeledyneS	
30				X	X					X	X			X		HS7541B-1	† HybridSys	
				X	X					X	X			X		HS7541C-1	HybridSys	50
				X						X	X			X		DAC1209	National	

(Continued)

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ± LSB	Settling Time ± ½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																	<b>(Cont'd)</b>		
12	1	1	30																
				x						x	x				x		DAC1219	National	(Cont'd)
				x						x	x				x		DAC1231	National	
		40 *		x	x	x	x			x	x				x	x	MP7622A	MicroPwr (3046)	
				x	x	x	x			x	x				x	x	MP7622J	MicroPwr (3046)	
				x	x	x	x			x	x				x	x	MP7622S	† MicroPwr (3046)	5
				x	x	x	x			x	x				x	x	SI7622A	Siliconix (3083)	
				x	x	x	x			x	x				x	x	SI7622S	† Siliconix (3083)	
				x	x	x	x			x	x				x	x	SI7722J	Siliconix	
		50 *		x	x	x	x			x	x				x		AD7541S	† AD (2825,2827,3351)	
				x	x	x	x			x	x				x		AD7541L	Intersil	10
				x	x	x	x			x	x				x		AD7541S	† Intersil	
		450		x	x	x	x			x	x				x		MP7623A	MicroPwr (3046)	
				x	x	x	x			x	x				x		MP7623J	MicroPwr (3046)	
				x	x	x	x			x	x				x		MP7623S	† MicroPwr (3046)	
				x	x					x	x						TDP75415	† TeledyneP	15
				x	x					x	x						TDP7541A	TeledyneP	
				x	x					x	x						TDP7541J	TeledyneP	
		1 *	500	x	x						x						MP1209A	MicroPwr	
				x	x						x						MP1209J	MicroPwr	
				x	x						x						MP1209S	† MicroPwr	20
				x	x						x					x	MP1231A	MicroPwr	
				x	x						x					x	MP1231J	MicroPwr	
				x	x						x					x	MP1231S	† MicroPwr	
				x	x						x					x	MP1232H	MicroPwr	
				x	x						x					x	MP1232R	† MicroPwr	25
				x	x						x					x	MP1232Z	MicroPwr	
		2	40	x	x					x	x					x	AD7542A	AD (2827)	
				x	x					x	x					x	AD7542J	AD (2827)	
				x	x					x	x					x	AD7542S	† AD (2827)	
				x	x					x						x	AD7543A	AD (2827)	30
				x	x					x						x	AD7543J	AD (2827)	
			450	x	x						x					x	MP7542S	† Harris	
				x	x						x					x	MP7542A	MicroPwr (3046)	
				x	x						x					x	MP7543A	MicroPwr (3046)	
				x	x						x					x	MP7543J	MicroPwr (3046)	35
				x	x						x					x	MP7543S	† MicroPwr (3046)	
			670	x	x						x					x	MP7542J	MicroPwr (3046)	
		2.5	—	x	x					x	x				x		DAC338B-12-1	HybridSys (2983)	
		5	450 *	x	x					x	x				x		HS9338-1	HybridSys (2983)	
		10	750			x	x	x		x					x		DAC-SL-11	† DDC	40
		20 *	500	x	x					x	x				x		DAC9377-16-5	HybridSys (2984)	
		30	—	x		x									x		M/DA2005	HyComp	
		35	150			x				x	x				x		AD371K	AD (2826)	
						x				x	x				x		AD371S	† AD (2826)	
		550 *	30														AD7240K	AD (2820,2825)	45
																	AD7240T	† AD (2820,2825)	
		1/2	0.5	330	x					x	x				x		HI5660-2	† Harris (2950)	
										x	x				x		HI5660-5	Harris (2950)	
										x	x				x		HI5660A-2	† Harris (2950)	
										x	x				x		HI5660A-S	Harris (2950)	50
		1 *	500	x	x						x						MP1208B	MicroPwr (3046)	
				x	x						x						MP1208K	MicroPwr (3046)	
				x	x						x						MP1208T	† MicroPwr (3046)	
				x	x						x					x	MP1230B	MicroPwr (3046)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linearity Error $\pm$ LSB	Settling Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
12	1/2	1*	500	x	x						x				x	MP1230K	MicroPwr (3046)	
				x	x						x				x	MP1230T	† MicroPwr (3046)	
		2	450	x	x						x				x	MP7542B	MicroPwr (3046)	
				x	x						x				x	MP7542T	† MicroPwr (3046)	
				x	x						x				x	MP7543B	MicroPwr (3046)	5
				x	x						x				x	MP7543K	MicroPwr (3046)	
				x	x						x				x	MP7543T	† MicroPwr (3046)	
			670	x	x						x				x	MP7542K	MicroPwr (3046)	
	1/4	1*	500	x	x						x					MP1210H	MicroPwr	
				x	x						x					MP1210R	† MicroPwr	10
				x	x						x					MP1210Z	MicroPwr	
	1 1/2	550*	30													AD7240J	AD (2820,2825)	
																AD7240S	† AD (2820,2825)	
	2	—	35		x			x	x	x						AD7545A	AD (2821,2827,2827)	
					x			x	x	x						AD7545J	AD (2821,2827,2827)	15
					x			x	x	x						AD7545S	† AD (2821,2827,2827)	
	0.04	—		x	x							x	x			DA4000	HyComp	
	0.05	—				x	x					x	x			ADH-030II-10	DDC	
	0.25			x	X	x	x			x	x			x	x	DAC-6012AC	Raytheon (3590)	
				x	X	x	x			x	x			x	x	DAC-6012AM	Raytheon (3590)	20
				x	X	x	x			x	x			x	x	DAC-6012C	Raytheon (3590)	
				x	X	x	x			x	x			x	x	DAC-6012M	Raytheon (3590)	
			375	x		x				x	x	x			x	DAC-312B	† PMI (3567)	
				x		x				x	x	x				DAC-312F	PMI (3567)	
	0.5	10*		x						x	x				x	DAC1220	† National	25
				x						x	x				x	DAC1220C	National	
		20*		x	x	x	x			x	x				x	MP7621Z	MicroPwr (3046)	
	0.5*	20*		x	x	x	x			x	x				x	AD7521L	AD (2825)	
				x	x	x	x			x	x				x	AD7521U	† AD (2825)	
				x	x	x	x			x	x				x	AD7521L	Intersil	30
				x	x	x	x			x	x				x	AD7521U	† Intersil	
				x	x	x	x			x	x				x	AD7531L	Intersil	
				x	x	x	x			x	x				x	MP7521L	MicroPwr (3046)	
				x	x	x	x			x	x				x	MP7521U	† MicroPwr (3046)	
				x	x	x	x			x	x				x	MP7531L	MicroPwr (3046)	35
	0.50	397		x	x	x		x		x	x	x			x	AM6012	Signetics (3605)	
	1	30		x						x	x				x	DAC1210	National	
				x						x	x				x	DAC1232	National	
	2	40*		x	x	x	x			x	x				x	MP7622H	MicroPwr (3046)	40
				x	x	x	x			x	x				x	MP7622R	† MicroPwr (3046)	
				x	x	x	x			x	x				x	MP7622Z	MicroPwr (3046)	
				x	x	x	x			x	x				x	SI7622H	Siliconix (3083)	
				x	x	x	x			x	x				x	SI7622R	† Siliconix (3083)	
				x	x	x	x			x	x				x	SI7622Z	Siliconix (3083)	
		375*		x	x	x				x	x				x	DAC3120B-0	† HybridSys	45
				x	x	x				x	x				x	DAC3120C-0	HybridSys	
	2*	—				x	x			x						DAC345I-10	HybridSys	
	2.5	—		x	x					x	x				x	DAC338B-12-0	HybridSys (2983)	
	5	450*		x	x					x	x				x	HS933B-0	HybridSys (2983)	
	5*	—				x	x			x						DAC345V-10	HybridSys	50
	20*	500		x	x					x	x				x	DAC9377-16-4	HybridSys (2984)	
	35*	—		x	x					x	x				x	DDC1250-10-1	DDC	
				x	x					x	x				x	DDC1250-10-3	DDC	

(Continued)

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error ±LSB	Setting Time ±½ LSB μS	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																	<b>(Cont'd)</b>		
12	2	500 *	24	x						x	x				x	AD7521L	National	(Cont'd)	
				x						x	x				x	AD7521U	† National		
	4	0.5 *	20 *	x	x	x	x			x	x				x	AD7521K	AD (2825)		
				x	x	x	x			x	x				x	AD7521T	† AD (2825)		
				x	x	x	x			x	x				x	AD7531K	AD (2825)	5	
				x	x	x	x			x	x				x	AD7521K	Intersil		
				x	x	x	x			x	x				x	AD7521T	† Intersil		
				x	x	x	x			x	x				x	AD7531K	Intersil		
				x	x	x	x			x	x				x	MP7521K	MicroPwr (3046)		
				x	x	x	x			x	x				x	MP7521T	† MicroPwr (3046)	10	
				x	x	x	x			x	x				x	MP7531K	MicroPwr (3046)		
		500 *	24	x						x	x				x	AD7521K	National		
				x						x	x				x	AD7521T	† National		
				x						x	x				x	DAC1221	† National		
				x						x	x				x	DAC1221C	National	15	
	8	0.5 *	20 *	x	x	x	x			x	x				x	AD7521J	AD (2825)		
				x	x	x	x			x	x				x	AD7521S	† AD (2825)		
				x	x	x	x			x	x				x	AD7531J	AD (2825)		
				x	x	x	x			x	x				x	AD7521J	Intersil		
				x	x	x	x			x	x				x	AD7521S	† Intersil	20	
				x	x	x	x			x	x				x	AD7531J	Intersil		
				x	x	x	x			x	x				x	MP7521J	MicroPwr (3046)		
				x	x	x	x			x	x				x	MP7521S	† MicroPwr (3046)		
				x	x	x	x			x	x				x	MP7531J	MicroPwr (3046)		
		500 *	24	x						x	x				x	AD7521J	National	25	
				x						x	x				x	AD7521S	† National		
				x						x	x				x	DAC1222	† National		
				x						x	x				x	DAC1222C	National		
<b>12 Plus Sign Dynamic Range (7-Bit plus sign format)</b>																			
		—	0.5							x						AM6072C	AMD		
										x						AM6072M	† AMD	30	
	1/2 step	0.5	192							x						AM6070AC	AMD		
										x						AM6070AM	† AMD		
										x	x	x		x		DAC-86E	PMI		
	1 step	0.5	192							x						AM6070C	AMD		
										x						AM6070M	† AMD	35	
										x	x	x		x		DAC-86C	PMI		
<b>12-3 Digit BCD</b>																			
	1/4	0.3	770 *			x				x	x		x			DAC-85C-CCD-I	Datel		
						x				x	x		x			DAC-87-CCD-I	† Datel (2861,2861)		
		0.3/3 *	925							x	x		x			AD8080/CCD	AD (2825)		
						x				x	x		x			AD8085/CCD	† AD (2825)	40	
						x				x	x		x			AD8085C/CCD	AD (2825)		
	3	770 *				x				x	x		x			DAC-85C-CCD-V	Datel		
						x				x	x		x			DAC-87-CCD-V	† Datel (2861,2861)		
		900											x	x		DAC-NK12DGC	Datel (2861)		
													x		x	DAC-NK12DMM	† Datel (2861)	45	
	3 *	1050								x			x			DAC-NZ12DGC	Datel (2861)		
										x			x			DAC-NZ12DMC	Datel (2861)		
										x			x			DAC-NZ12DMM	† Datel (2861)		
	1/2	0.3	1200								x		x			DAC80-CCD-I	Burr-Brown (2848)		
											x		x			DAC80Z-CCD-I	Burr-Brown (2848)	50	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

(Continued)

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error ±LSB	Settling Time ±1/2 LSB μs	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
12-3 Digit BCD																	(Cont'd)	
	0.3/3 *	800								x			x			DAC85C/CCD	Burr-Brown (2848)	
		850 *								x			x			DA85C-CCD	MicroNet	
										x			x			DAC85-CCD †	MicroNet (3044)	
	0.5	30								x	x		x			DAC-HA12DC-1	Datel (2860)	5
										x	x		x			DAC-HA12DM-1 †	Datel (2860)	
	0.5 *	0.05								x	x		x			DAC-HA12DC	Datel (2860)	
										x	x		x			DAC-HA12DM †	Datel (2860)	
	15	300		x	x					x	x		x			DAC349B-3D †	HybridSys	
				x	x					x	x		x			DAC349C-3D	HybridSys	
				x						x	x		x			DAC9349-3D	HybridSys	10
	1/10	1.5 *	465							x	x			x		AD562K/BCD	AD (2827,2829)	
										x	x			x		AD562S/BCD †	AD (2827,2829)	
		475								x	x		x			AD563J/BCD	AD (2826)	
										x	x		x			AD563K/BCD	AD (2826)	
										x	x		x			AD563S/BCD †	AD (2826)	15
										x	x		x			AD563T/BCD †	AD (2826)	
12-4 Digit BCD																		
	1/2	20 *	500	x	x					x	x		x		x	DAC9377-4D	HybridSys (2984)	
13	1	0.1	980				x			x			x			2615-12	DDC	
		1.6	1100				x			x			x			SDAC-12	DDC	
	1/2	10	360	x	x					x	x					MP8526B	MicroPwr (3046)	20
				x	x					x	x					MP8526S †	MicroPwr (3046)	
	1/4	10	360	x	x					x	x					MP8526C	MicroPwr (3046)	
				x	x					x	x					MP8526T †	MicroPwr (3046)	
	2	0.1	980				x			x			x			2615-11	DDC	
		1.6	1100				x			x			x			SDAC-11	DDC	25
	4	0.1	980				x			x			x			2615-10	DDC	
		1.6	1100				x			x			x			SDAC-10	DDC	
13 (3 Device Set)																		
	1	1.8	3475					x		x			x			DDAC-12	DDC	
	2	1.8	3475					x		x			x			DDAC-11	DDC	
	4	1.8	3475					x		x			x			DDAC-10	DDC	30
14	1/2			x	x					x	x			x		DAC9331-14	HybridSys (2984)	
	2	30 *		x	x					x	x			x		HS3140B-4 †	HybridSys (2985)	
				x	x					x	x			x		HS3140C-4	HybridSys (2985)	
	3			x				x		x	x			x	x	ICL7134	Intersil (3001)	
		30		x	x					x	x			x		DAC331B-14 †	HybridSys	35
				x	x					x	x			x		DAC331C-14	HybridSys	
	15	1135		x	x			x		x			x			MP1814	Analogic	
		1950		x				x		x			x		x	MP1914A	Analogic	
		2850		x				x		x			x		x	MP1914TC	Analogic	
		200 kHz								x						HDD-1409	AD (2823)	40
	1	0.5	0.05	x	x					x	x			x		DAC-HA14BC	Datel (2860)	
				x	x					x	x			x		DAC-HA14BM †	Datel (2860)	
		30		x	x					x	x			x		DAC-HA14BC-1	Datel (2860)	
				x	x					x	x			x		DAC-HA14BM-1	Datel (2860)	
	2	30 *		x	x					x	x			x	x	MP3140B-4 †	MicroPwr (3046)	45
				x	x					x	x			x	x	MP3140C-4	MicroPwr (3046)	
		375 *		x	x	x				x	x			x		HS3140B-3	HybridSys (2985)	
				x	x	x				x	x			x		HS3140C-3	HybridSys (2985)	
	2	2	20	x	x					x	x			x	x	MP7614K	MicroPwr (3046)	
				x	x					x	x			x	x	MP7614T †	MicroPwr (3046)	50
		30 *		x	x					x	x			x	x	MP3140B-3 †	MicroPwr (3046)	
				x	x					x	x			x	x	MP3140C-3	MicroPwr (3046)	

(Continued)

Bin.—Binary  
Off.—Offset  
Compl.—Complementary  
Int Ref.—Internal Reference  
CTC—Compl. 2's Compl.  
Mult.—Multiplying

**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error $\pm 1/2$ LSB $\pm$ LSB	Settling Time $\mu$ s	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																	<b>(Cont'd)</b>	
14	2	20	900	x						x					x	DAC-U12-1	† DDC	(Cont'd)
	4	2	20	x	x					x	x				x	MP7614J	MicroPwr (3046)	
				x	x					x	x				x	MP7614S	† MicroPwr (3046)	
		20	900	x	x					x					x	DAC-U11-1	† DDC	
15	1/2	17	1950	x				x		x			x		x	MP1915A	Analogic	5
			2850	x				x		x			x		x	MP1915TC	Analogic	
16	1/8	25	1425	x				x		x			x		x	MP8116	Analogic	
	1/4	3.5	750		x			x	x	x	x		x			MP1936	Analogic	
		5	500		x			x	x	x	x		x			MP1926A	Analogic	
		50	500		x			x	x	x	x		x			MP1926S	Analogic	10
	1/2	2 *	60	x	x					x	x				x	DAC370B-16	† HybridSys	
				x	x					x	x				x	DAC370C-16	HybridSys	
				x	x					x	x				x	DAC9331-16-6	HybridSys (2984)	
				x	x					x	x				x	MP9331-6	MicroPwr (3046)	
	5 *	50 *		x	x					x	x				x	AD7546B	AD (2827,2829)	15
				x	x					x	x				x	AD7546K	AD (2827,2829)	
	20		1950	x				x		x			x		x	MP1916A	Analogic	
			2850	x				x		x			x		x	MP1916TC	Analogic	
	20 *	500		x	x					x	x		x		x	DAC9377-16-6	HybridSys (2984)	
				x	x					x	x		x		x	MP9377-16	MicroPwr (3046)	20
	50	1700				x	x			x	x		x			DAC736K	Burr-Brown (2848)	
1	1	465		x	x			x		x	x					HIDAC16B	Harris (2946)	
	2 *	60		x	x					x	x				x	DAC9331-16-5	HybridSys (2984)	
				x	x					x	x				x	MP9331-5	MicroPwr (3046)	
	5 *	50 *		x	x					x	x				x	AD7546A	AD (2827,2829)	25
				x	x					x	x				x	AD7546J	AD (2827,2829)	
	20	450		x	x					x	x				x	MP9377-16-5	MicroPwr (3046)	
	50	2000				x	x			x	x		x		x	DAC736J	Burr-Brown (2848)	
						x	x			x	x		x		x	DAC73J	Burr-Brown (2848)	
	1/2	20	450	x	x					x					x	MP9377-16-6	MicroPwr (3046)	30
2	0.3	940				x				x	x		x			DAC700B	Burr-Brown (2848)	
						x				x	x		x			DAC700K	Burr-Brown (2848)	
							x	x		x	x		x			DAC702B	Burr-Brown (2848)	
							x	x		x	x		x			DAC702K	Burr-Brown (2848)	
	1	-				x				x			x			DAC-71-COB-1	Datel	35
						x				x			x			DAC-71-CSB-1	Datel	
						x				x			x			DAC-72C-COB-1	Datel	
						x				x			x			DAC-72C-CSB-1	Datel	
		465		x	x			x		x	x					HIDAC16C	Harris (2946)	
		1175					x	x		x			x			DAC72C-COB-1	Burr-Brown (2848)	40
						x				x			x			DAC72C-CSB-1	Burr-Brown (2848)	
		1225					x			x			x			ADDAC71-COB-1	AD (2827,2829)	
						x				x			x			ADDAC71-CSB-1	AD (2827,2829)	
							x			x			x			ADDAC72-COB-1	AD (2827,2829)	
						x				x			x			ADDAC72-CSB-1	AD (2827,2829)	45
	2	20		x	x					x	x				x	MP7616T	† MicroPwr (3046)	
		30		x	x					x	x				x	HS3160B-4	† HybridSys (2985)	
				x	x					x	x				x	HS3160C-4	HybridSys (2985)	
	2 *	60		x	x					x	x				x	DAC9331-16-4	HybridSys (2984)	
				x	x					x	x				x	MP9331-4	MicroPwr (3046)	50
	3							x		x	x				x	ICL7145	Intersil (3007)	
	4	790				x				x	x		x			DAC701B	Burr-Brown (2848)	
						x				x	x		x			DAC701K	Burr-Brown (2848)	
							x	x		x	x		x			DAC703B	Burr-Brown (2848)	
							x	x		x	x		x			DAC703K	Burr-Brown (2848)	55

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Digital to Analog Converters (Cont'd)

Bits Res.	Linear-ity Error $\pm$ LSB	Settling Time $\mu$ S	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Off. Bin. Input	Compl. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line
<b>D/A Converters</b>																		<b>(Cont'd)</b>	
16	2	5	—					X			X			X			DAC-71-COB-V	Datel	(Cont'd)
								X			X			X			DAC-71-CSB-V	Datel	
								X			X			X			DAC-72C-COB-V	Datel	
								X			X			X			DAC-72C-CSB-V	Datel	
		1225						X			X			X			ADDAC71-COB-V	AD (2827,2829)	5
								X			X			X			ADDAC71-CSB-V	AD (2827,2829)	
								X			X			X			ADDAC72-COB-V	AD (2827,2829)	
								X			X			X			ADDAC72-CSB-V	AD (2827,2829)	
	10	1175						X			X			X			DAC71-CSB-V	Burr-Brown (2848)	10
								X	X		X			X			DAC72-COB-V	Burr-Brown (2848)	
								X			X			X			DAC72-CSB-V	Burr-Brown (2848)	
								X	X		X			X			DAC72C-COB-V	Burr-Brown (2848)	
								X			X			X			DAC72C-CSB-V	Burr-Brown (2848)	
	20	450		X	X							X			X	X	MP9377-16-4	MicroPwr (3046)	15
	35	1170						X	X		X			X			DAC-HP16BMC	Datel (2861)	
								X	X		X			X			DAC-HP16BMC-1	Datel (2861)	
								X	X		X			X			DAC-HP16BMM-1	Datel (2861)	
								X	X		X			X			DAC-HP16BMM-1	† Datel (2861)	
	50/100 *	575 *						X			X			X			DAC70-CSB-I	† Burr-Brown (2848)	20
								X			X			X			DAC70C-CSB-I	Burr-Brown (2848)	
3	50/100 *	575 *						X			X			X			DAC70-COB-I	† Burr-Brown (2848)	
								X			X			X			DAC70C-COB-I	Burr-Brown (2848)	
4	2	—		X	X						X	X			X		MP7616L	MicroPwr (3046)	25
		20		X	X						X	X			X	X	MP7616J	MicroPwr (3046)	
				X	X						X	X			X	X	MP7616S	† MicroPwr (3046)	
	3 *	425						X	X		X	X					PCM52	Burr-Brown (2849)	
								X	X		X	X					PCM53	Burr-Brown (2849)	
8	2	—		X	X						X	X			X		MP7616K	MicroPwr (3046)	30
		30		X	X						X	X			X		HS3160B-3	† HybridSys (2985)	
				X	X						X	X			X		HS3160C-3	HybridSys (2985)	
16-4 Digit BCD	1/2	15	1170								X			X			DAC-HP16DGC	Datel (2861)	35
											X			X			DAC-HP16DMC	Datel (2861)	
											X			X			DAC-HP16DMM	† Datel (2861)	
	50/100 *	575 *									X			X			DAC70-CCD-I	† Burr-Brown (2848)	
											X			X			DAC70C-CCD-I	Burr-Brown (2848)	
2	1	—						X			X			X			DAC-71-CCD-I	Datel	40
								X			X			X			DAC-72C-CCD-I	Datel	
		1175									X			X			DAC72C-CCD-I	Burr-Brown (2848)	
	5	—						X			X			X			DAC-71-CCD-V	Datel	
								X			X			X			DAC-72C-CCD-V	Datel	
	10	1175									X			X			DAC71-CCB-V	Burr-Brown (2848)	
								X			X			X			DAC71-COB-V	Burr-Brown (2848)	
											X			X			DAC72-CCD-V	Burr-Brown (2848)	
											X			X			DAC72C-CCD-V	Burr-Brown (2848)	
18	1	20	60	X	X						X	X		X		X	DAC370B-18	† HybridSys	45
				X	X						X	X		X		X	DAC370C-18	HybridSys	
				X	X						X	X		X		X	MP370B-18	† MicroPwr (3046)	
				X	X						X	X		X		X	MP370C-18	MicroPwr (3046)	
		500		X	X						X	X		X			DAC377-18	† HybridSys (2984)	50
				X	X						X	X		X		X	DAC377B-18	† HybridSys (2984)	
				X	X						X	X		X		X	DAC377C-18	HybridSys (2984)	

Bin.—Binary  
Off.—Offset

Magn.—Magnitude

Compl.—Complementary

Int Ref.—Internal Reference

CTC—Compl. 2's Compl.  
Mult.—Multiplying



**IC MASTER**

**INTERFACE-Digital to Analog Converters (Cont'd)**

Bits Res.	Linear-ity Error $\pm$ LSB	Settling Time $\pm$ 1/2 LSB $\mu$ S	Power Dis. mW (max.)	Bin. Input	Off. Bin. Input	Compl. Bin. Input	Compl. Off. Bin. Input	CTC or 2's Compl. Input	Sign. Magn. Input	TTL Logic	CMOS Logic	ECL Logic	Int. Ref.	Mult.	Latches	Device	Source	Line	
<b>D/A Converters</b>																<b>(Cont'd)</b>			
18	1	20	500	X	X					X	X						X		
				X	X					X	X						X		
																	<b>MP377B-18</b>	<b>†MicroPwr</b>	<b>(3046)</b>
																	<b>MP377C-18</b>	<b>MicroPwr</b>	<b>(3046)</b>

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Display Drivers

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line	
<b>Display Drivers</b>				LCD Driver, Multiplexed					BCD to 7-Segment Decoder/Driver, Active High, Open Collector			
Bargraph Fluorescent	XR2276	Exar		PCF8576	Signetics			DM74LS49	National			
Bargraph LED	LM3914	National		LCD Driver, 32/64 Segment	PCF8577	Signetics	55	SN54249	† TI	(1002)		
	LM3915	National		LCD Graphics Driver	SED1500	SMOS		SN5449	† TI	(935)		
	AN6875	Panasonic		LED Bar Display Driver, 10 LEDs	TA7612A	Toshiba		SN54LS249	† TI	(1002)		
	LB1405	Sanyo	5	LED, Cascadable	SDA2014	Siemens		SN54LS49	† TI	(935)	105	
	LB1409	Sanyo		LED Display Driver, 5 LEDs	TA7654	Toshiba	10	SN74249	TI	(1002)		
	LB1415	Sanyo		TA7655	Toshiba		SN7449	TI	(935)			
	LB1416	Sanyo		LED Display Driver, 33 Outputs, 15 mA Sink Capability	MM5486	National		SN74LS249	TI	(1002)		
	LB1419	Sanyo		LED Display/Interface (drives 7, 14, 16-segments for linear displays, bar graphs)	SAA1060	Signetics	15	SN74LS49	TI	(935)		
	LB1426	Sanyo		SAA1061	Signetics							
	LB1436	Sanyo		LED Driver System, 8 Decade, 8x8 Memory, Decoder (versions either hardwire or processor controlled)	ICM7218A	Intersil (786)						
	UAA170	Siemens		ICM7218B	Intersil (786)		65					
	UAA180	Siemens		ICM7218C	Intersil (786)							
	U237B	Telefunken		ICM7218D	Intersil (786)							
	U244	Telefunken		ICM7218E	Intersil (786)							
	U247B	Telefunken		LED, Static	U180M	Telefunken						
	U254	Telefunken		U3080M	Telefunken		70					
	U257B	Telefunken		Multiplexed LCD Driver, Master and Slave	MC145000	Motorola						
	U267B	Telefunken		MC145001	Motorola							
	TSC9403	Toshiba	20	MM58201	National							
	TSC9404	Toshiba		Segment Driver for Gas Discharge Displays	DI232	Dionics						
Bargraph LED or Vacuum Fluorescent (module)	AW580	AWI		DI242	Dionics							
Bargraph or LED Dot Display Generators	XR2277	Exar (3375)		Universal (LED, LCD, or vacuum fluorescent)	S2809	AMI						
	XR2278	Exar (3375)		S2809	AMI							
	XR2279	Exar (3375)	25	BCD (hexadecimal) to 7-Segment Decoder/Driver with Latch Active Low, Open Collector	9370C	Fairchild						
Bargraph VF	HA12010	Hitachi		9370C	Fairchild							
	HA12011	Hitachi		BCD (hexadecimal) to 7-Segment Latch/Decoder/Driver (CMOS with bipolar output)	MC14495	Motorola						
	LM3916	National		UCN-4805A	Sprague							
	LB1470	Sanyo		UCN-4806A	Sprague							
Bargraph 18-Element LCD Driver	HEF9754V	Signetics	30	BCD Plus 1 to 7-Segment	LM1017	National						
Clock Driver, Dual	DS3671	National		U143	Telefunken							
Display Controllers and Keyboard Interface: See Microprocessors-General Purpose				BCD to 7-Segment Decoder/Driver	MC14547BA	† Motorola						
Driver, 16-Channel Electroluminescent Display Column Driver	HV1	Supertex		MC14547BC	Motorola		85					
Driver, 16-Channel Electroluminescent Display Row Driver	HV2	Supertex		BCD to 7-Segment Decoder/Driver, Active High, Open Collector	5449	† Fairchild						
Electroluminescent Even Column Driver, 32 Outputs	SN75554	TI	35	54LS249	† Fairchild							
Electroluminescent Even Row Driver, 32 Outputs	SN75552	TI		54LS49	† Fairchild							
Electroluminescent Odd Column Driver, 32 Outputs	SN75553	TI		74LS249	Fairchild							
Electroluminescent Odd Row Driver, 32 Outputs	SN75551	TI		74LS49	Fairchild							
Lamp Driver	CSR301	TeledyneC	40	HD74LS249	Hitachi							
LCD Column Driver	ICM7261	Intersil (3020)		MC7449	Motorola							
LCD Display Interface (drives 7 to 20-segment linear displays)	SAA1062	Signetics		SN54LS249	Motorola							
LCD Dot Matrix Driver	HLCD0488	Hughes		SN54LS49	† Motorola							
	HLCD0515	Hughes		SN74LS249	† Motorola							
	HLCD0538A	Hughes	45	SN74LS49	Motorola							
	HLCD0539A	Hughes		DM54LS249	† National							
	HLCD0540	Hughes		DM54LS49	† National							
	HLCD0548	Hughes		DM74LS249	National		100					
	HLCD0550	Hughes										
	HLCD0551	Hughes	50									
	HLCD0607	Hughes										
	SED1100	SMOS										
	SED1300	SMOS										

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## INTERFACE-Display Drivers (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line	
<b>Display Drivers (Cont'd)</b>												
BCD to 7-Segment Decoder/Driver, Active Low, Open Collector, 15 V Output (Cont'd)				BCD to 7-Segment Latch/Decoder/Driver, for Liquid Crystal Displays				4/5-Digit Fluorescent Display Driver				
	SN74LS47	Motorola		<b>CD4543BD</b>	† RCA	(839)		MM5474	National		105	
	DM5447A	† National		<b>CD4543BE</b>	RCA	(839)		MM5476	National			
	DM54LS247	† National		BCD to 7-Segment Latch/Decoder/Driver, Output 1-16					MM5477	National		
	DM54LS47	† National		SAB3211	Siemens			MM5478	National			
	DM7447A	National	5	BCD to 7-Segment Latch/Decoder/Driver (ripple blanking)				4/5-Digit (serial data input)				
	DM74LS247	National		MC14544BA	† Motorola		60	MM5450	National			
	DM74LS47	National		MC14544BC	Motorola			MM5451	National			
	383A/C	TeledyneS		BCD to 7-Segment Latch/Decoder/Driver (strobed latch), for Liquid Crystal Displays				4/5-Digit Vacuum Fluorescent				
	383B/M	† TeledyneS		<b>LS7100</b>	LSIComp	(803)		MM5445	National		110	
	<b>SN54247</b>	† TI	10	MC14543BA	† Motorola			MM5446	National			
	<b>SN5447A</b>	† TI	(935)	MC14543BC	Motorola			MM5447	National			
	<b>SN54LS247</b>	† TI	(1001)	<b>CD4056B</b>	† RCA	(839)	65	MM5448	National			
	<b>SN54LS247A</b>	† TI	(1001)	<b>CD4056BE</b>	RCA	(839)		SAA 1063	Signetics			
	<b>SN54LS47</b>	† TI	(935)	883/4543B	† SSS			4x5 Matrix LED Decoder/Driver, Interface to CPU				
	<b>SN74247</b>	TI	(1001)	SCL4543B	SSS			MN 1205F	Panasonic		115	
	<b>SN7447A</b>	TI	(935)	TC4056B	Toshiba			4x8 Matrix LED Decoder/Driver, Interface to MPU				
	SN74L47	TI		BCD to 7-Segment LED Decoder/Driver					MN 1205A	Panasonic		
	<b>SN74LS247</b>	TI	(1001)	<b>CA3168</b>	RCA	(3599)	70	4-Bit Multiplexed LCD Driver (4x7)				
	<b>SN74LS47</b>	TI	(935)	BCD to 7-Segment LED Decoder/Driver, Constant Current					HLCD7211-1	Hughes		
BCD to 7-Segment Decoder/Driver, Active Low, Open Collector, 30 V Output				BCD to 7-Segment LED Decoder/Driver with Latch, Output 0-9, —, E,H,L,P				4-Channel Incandescent Lamp Driver				
	5446	† Fairchild	20	DS8857	National			CLD4	TeledyneC			
	7446	Fairchild		NE587	Signetics			4-Channel Plasma				
	9317CC	Fairchild		BCD to 7-Segment LED Decoder/Driver with Latch, Output 0-9, —, E,H,L,P					XR2284	Exar	120	
	9317CM	† Fairchild		9374C	Fairchild			4-Character, 18-Segment Triplexed LCD Decoder/Driver				
	HD7446A	Hitachi		BCD to 7-Segment LED Driver Adjustable Current					<b>ICM7233</b>	Intersil	(798)	
	MC7446	Motorola	25	DS8858	National			4-Digit Driver (Quad Power Stroke)				
	DM5446A	† National		BCD to 7-Segment LED Latch/Decoder/Driver, with Ripple Blanking					HD6600-2	† Harris		
	DM7446A	National		F4734BC	Fairchild		75	HD6600-5	Harris			
	<b>SN54246</b>	† TI	(1001)	F4734BM	† Fairchild			HD6600	National			
	<b>SN5446A</b>	† TI	(935)	BCD-to-Decimal Decoder/Driver (for lamps)				4-Digit Fluorescent				
	<b>SN54LS46</b>	† TI	(935)	380A/C	TeledyneS			ICM7235	Intersil		125	
	<b>SN74246</b>	TI	(1001)	380B/M	† TeledyneS			ICM7235A	Intersil			
	<b>SN7446A</b>	TI	(935)	381A/C	TeledyneS			ICM7235AM	Intersil			
	<b>SN74LS46</b>	TI	(935)	381B/M	† TeledyneS			ICM7235M	Intersil			
BCD to 7-Segment Decoder/Driver, for Fluorescent Displays				BCD-to-Decimal Decoder/Driver (nixie driver)				4-Digit Gas Discharge Display Anode Driver				
	CS250	Cherry	35	7441	Fairchild			DI500	Dionics		130	
	CS250-1	Cherry		DM5441A	† National			DI502	Dionics			
BCD to 7-Segment Decoder/Driver, for Liquid Crystal Displays				BCD-to-Decimal Decoder/Driver with Blanking (for cold cathode indicator tubes)				4-Digit LCD Driver				
	<b>CD4055B</b>	† RCA	(839)	DM7441A	National		85	<b>ICM7211AM</b>	Intersil	(779)		
	<b>CD4055BE</b>	RCA	(839)	382A/C	TeledyneS			<b>ICM7211M</b>	Intersil	(779)		
	TC4055B	Toshiba		382B/M	† TeledyneS			TSC7211AM	TeledyneS			
BCD to 7-Segment Decoder/Driver, 2-Digit, Direct Driver for Common Anode LED Displays				BCD-to-Decimal Decoder/Driver with Blanking (for cold cathode indicator tubes)				4-Digit LED Driver				
	DS8669	National		54141	† Fairchild			SDA2004	Siemens			
BCD to 7-Segment Latch/Decoder/Driver				Hex TTL to LED Bulb Driver, with Latch				4-Digit LED Driver, Multiplexed BCD or Binary to 7-Segment Decoder/Driver				
	CS260	Cherry	40	DS8859	National			<b>ICM7212</b>	Intersil	(779)	135	
BCD to 7-Segment Latch/Decoder/Driver (CMOS with bipolar output)				2-Digit, 7-Segment Decoder/Driver Interfaces to CPU				4-Digit LED Driver, BCD or Binary to 7-Segment Decoder/Driver, Data and Digit Select Code Latches for IP Interface				
	F4511BC	Fairchild		MN 1205E	Panasonic		100	<b>ICM7212A</b>	Intersil	(779)		
	F4511BM	† Fairchild		MN 1205P	Panasonic			<b>ICM7212M</b>	Intersil	(779)		
	HD14511B	Hitachi		MN 1205Q	Panasonic			UDN-7183A	Sprague			
	MC14511BA	† Motorola		Quad AC Plasma Display Axis Driver					UDN-7184A	Sprague		140
	MC14511BC	Motorola	45	SN55426B	† TI			UDN-7186A	Sprague			
	MC14513BA	† Motorola		SN55427B	† TI		95	UHP-482	Sprague			
	MC14513BC	Motorola		SN75426B	TI			<b>TSC701AM</b>	TeledyneS	(3093)		
	CD4511BC	National		SN75427B	TI			TSC7212A	TeledyneS			
	CD4511BM	† National		Hex TTL to LED Bulb Driver, with Latch					TSC7212AM	TeledyneS		145
	<b>CD4511B</b>	† RCA	(839)	DS8859	National			4-Digit Liquid Crystal, Multiplexed BCD to LCD Decoder/Driver, AC Drive				
	<b>CD4511BE</b>	RCA	(839)	DS8869	National			<b>ICM7211</b>	Intersil	(779)		
	883/4511B	† SSS		2-Digit, 7-Segment Decoder/Driver Interfaces to CPU					<b>ICM7211A</b>	Intersil	(779)	
	SCL4511B	SSS		MN 1205E	Panasonic			DF412	Siliconix			
	CM4511B	† Solitron		MN 1205P	Panasonic			SCL25411	SSS			
	CM4511BE	Solitron	55	MN 1205Q	Panasonic			4-Digit/Segment Fluorescent				
BCD to 7-Segment Latch/Decoder/Driver for Common Cathode LED Displays				3 1/2-Digit Liquid Crystal Clock				4-Digit/Segment Fluorescent				
	NE589	Signetics		C1200	LSIComp			DI503	Dionics		150	
								DI504	Dionics			

† Military Temperature Range (–55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE-Display Drivers (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Display Drivers (Cont'd)</b>											
4-Digit Select Microprocessor Interface (4x7)	HLCD7211-3 HLCD7211-4	Hughes Hughes		6-Digit MOS to LED Cathode Driver	55492A 75492 75492A MC75492	† Fairchild Fairchild Fairchild Motorola		8-Digit MOS to LED Cathode Driver	DS8863 DS8871 DS8963	National National National	
4-Digit (stores segment and address data, drives 7-8 segment digits)	MM74C911	National			DS75492 DS75494 DS8870 DS8877 PSS5494	National National National National † National	50	8-Digit/Segment Fluorescent	DI513 DI514 DI514A XR6118 XR6128 NE594 SA594	Dionics Dionics Dionics Exar Exar Signetics Signetics	100
4-Digit, 17-Segment Alpha-Numeric with Memory, Decoder and LED Drivers	MM74C956 NSM4307 NSM4507	National National National	5		SN75492 SN75492A SN75494	TI TI TI	55		SG6118 UDN-6118 UDN-6128 UDN-6138 UDN-6148	SiliconG Sprague Sprague Sprague Sprague	105
4-Digit, 7-Segment LCD Decoder/Driver	TSC7211A	TeledyneS		6-Digit/Segment Fluorescent	DI508 DI509 UDN-6116 UDN-6126	Dionics Dionics Sprague Sprague	60	8-Digit, 7-Segment Triplexed LCD Decoder/Driver	ICM7231	Intersil	(798)
4-Digit, 7-Segment LED Decoder/Driver	TSC700A	TeledyneS (3092)						8-Line Fluorescent Driver	MSL912R MSL915R MSL916R MSL917R	OKI OKI OKI OKI	(4118) (4118) (4118) (4118)
4-Digit/8-Segment Fluorescent	COP470	National		6-Digit BCD (stores segment and address data, drives 7-8 segment digits)	MM74C912	National	10				
4-Segment Liquid Crystal	CD4054B TC4054B	RCA Toshiba	(839)	6-Digit Hex (stores segment and address data, drives 7 segments)	MM74C917	National	15	8-Output Digit-Scan Counter/Decoder for Cold-Cathode Counter Tubes	MSL9510R MSL9511R	OKI OKI	(4118) (4118)
4-Segment MOS to LED Anode Driver	75491 MC75491 DS55493 DS75491 DS75493 SN75491 SN75491A	Fairchild Motorola † National National National TI TI		6-Digit, 7-Segment LCD Driver with Decimal Points, or Three 16-Segment Characters. 48-Stage Shift Register, 48-Bit Data Latch and 48-Segment Driver	MSM5219	OKI	65	8-Segment Gas Discharge Display Cathode Driver	DI210 DI2210 DI230 DI240 DI300 DI302 MC3491 MC3492 DS7889 DS8889 UDN-7180A	Dionics Dionics Dionics Dionics Dionics Dionics Motorola Motorola † National National National	120 125
5x7 Dot Matrix LCD/Console Controller	CY300	Cybernetic		7-Digit Gas Discharge Display Anode Driver	MC3490 MC3494	Motorola Motorola					
5x8 Dot Matrix Multiplexed LCD. On-board memory allows ASCII code and special symbols display without refresh circuitry. (2-chip set)	PE7901 PE7902	Polycore Polycore	20	7-Digit MOS to Gas Discharge	XR2272	Exar					
5x12 Dot Matrix (up to 80 characters by cascading)	10938 10939	Rockwell Rockwell		7-Digit MOS to LED Cathode Driver	SN75497	TI					
5-Character, 18-Segment Triplexed LCD Decoder/Driver	ICM7234	Intersil (798)		7-Digit/Segment MOS to Fluorescent	XR2271	Exar	70	7-Line Dot Matrix or Segmented	SN75581	TI	
5-Digit Gas Discharge Display Anode Driver	UHD-490 UHP-490	† Sprague Sprague	25	7-Line Dot Matrix or Segmented	SN75581	TI		7-Segment Gas Discharge Display Cathode Driver	DS8885 8885 UHP-481 SN75584A	National National Signetics Sprague TI	130
5-Line Plasma Display Axis Driver	DI5140 DI5180 DI5240 DI5280	Dionics Dionics Dionics Dionics	30	7-Segment Gas Discharge Display Cathode Driver, with BCD Decoder	DS7880 DS8880 DS8884A DS8980 DS8880 SN75480	† National National National National National Signetics TI					
5-Segment Gas Discharge Display Cathode Driver	UHP-480	Sprague		7-Segment to BCD Converter/Driver	MM54C915 MM74C915	† National National		8-Segment MOS to LED Anode Driver	MC676 DS8867	Motorola National	
5-Segment MOS to LED Anode Driver	DS8861	National		8-Bit Parallel In/Parallel Out Fluorescent (for μP systems)	UCN-4815A UCS-4815H	Sprague † Sprague	85	9-Digit MOS to LED Cathode Driver	DS8872 DS8920 DS8973 DS8975 SN75498	National National National National TI	135
5-Step Logarithmic Dual LED Driver	TA7666 TA7667	Toshiba Toshiba		8-Channel Plasma	XR2288	Exar		9-Digit MOS to LED Cathode Driver, with Low Battery Indicator	DS8864 DS8873	National National	
6-Digit Gas Discharge Display Anode Driver	DI505 DI507 DI603A DI604A DI605A DS8891 8891 UDN-6164A UHD-491 UHP-491 UHP-495	Dionics Dionics Dionics Dionics Dionics National National Sprague † Sprague Sprague Sprague	35 40 45	8-Character, 14/16 Segment LED Display Driver	ICM7243	Intersil (3017)		9-Digit MOS to LED Cathode Driver with Shift Register Decoding	DS8874	National	
6-Digit LED Driver (low voltage)	DS8646	National		8-Digit Gas Discharge Display Anode Driver	DI510 DI512 DI803A DI804A DI805A DS8887 DS8897 UDN-8184A	Dionics Dionics Dionics Dionics Dionics National National Sprague	90 95	9-Line, 30 LED	U1096B	Telefunken	
								9-Segment LED Driver (low voltage)	DS8647 DS8648	National National	140
								10-Bit High-Voltage, High-Current	S4534	AMI (2811)	
								10-Bit Serial In/Parallel Out Fluorescent (for μP systems)	UCN-4810A UCN4810A	Sprague TI	
								10-Digit, 7-Segment Triplexed LCD Decoder/Driver	ICM7232	Intersil (798)	145
								11-Segment LED Display Drivers	MM5485	National	
								12-Line Vacuum Fluorescent	SN75512A SN75513A SN75514	TI TI TI	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## INTERFACE-Display Drivers (Cont'd)

Function	Device	Source	Line
<b>Display Drivers (Cont'd)</b>			
14-Digit Decoder/Driver	DS8665	National	
	DS8666	National	
14-Digit Decoder/Driver, with Low Battery Indicator	DS7664	† National	
	DS8664	National	
14-Segment Decoder/Driver, Interface to CPU	MN 1205H	Panasonic	5
16-Line Fluorescent	DS8881	National	
	8881	Signetics	
16-Segment LED Display Drivers	MM5484	National	
18-Segment Alphanumeric	AC5947	TI	
18-Segment, 16-Character Alphanumeric (30, 35 and 40 volt versions)	10937	Rockwell	10
30-Bit LCD Driver/Register	MD4330B	Mitel	
32 Segment LCD	MM5452	National	
	MM5453	National	
	PCE2112	Signetics	
32-Bit High-Voltage	<b>S4521</b>	<b>AMI</b>	<b>(2810)</b>
	<b>S4535</b>	<b>AMI</b>	<b>(2812)</b>
32-Bit LCD	MM58438	National	15
32-Line AC Plasma Display Axis Driver	SN75500A	TI	
	SN75501C	TI	
32-Line Vacuum Fluorescent	SN75518	TI	20
32-Segment LCD Controller/Driver	μPD7255	NEC	
32-Segment LCD Driver	MM5483	National	
40-Segment LCD Duplex	PCE2100	Signetics	
60-Segment LCD Duplex	PCE2110	Signetics	
64-Segment LCD Duplex	PCF2111	Signetics	25

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE-Error Checking Circuits

Function	Max. Serial Data Rate, MHz	Supply Voltage, V	Device	Source	Line	
<b>Error Checking Circuits</b>						
CRC Generator/Checker	2 *	5	COM8004	SMC	5	
	3	5	MC8500	Motorola		
			<b>SSI8500</b>	<b>SiliconSys (4208)</b>		
	10	5	9401C	Fairchild		
			9401M	† Fairchild	10	
			8X01A	Signetics		
	12	5	54F401	† Fairchild		
			74F401	Fairchild		
CRC Generator/Checker (expandable)	20	5	9401M	† Fairchild	15	
			<b>N9401</b>	<b>Signetics (1656)</b>		
			<b>S9401</b>	<b>† Signetics (1656)</b>		
			54F402	Fairchild (738)		
Deskew-Queue Register	3 *	5	54F492	Fairchild	20	
			<b>74F402</b>	<b>Fairchild (738)</b>		
			74F492	Fairchild		
MC8520	Motorola	15				
<b>SSI8520</b>	<b>SiliconSys (4208)</b>					
Error Detection and Support Logic, for Winchester Disk System	5	5	<b>WD1014</b>	<b>Western (1724)</b>		
Error Detection/Correction Circuit	—	5	8206	Intel		
Error Detection/Correction Circuit (ECL)	—	-5.2	<b>WB8206</b>	<b>Western (4260)</b>	20	
			MC10163	Motorola		
			MC10193	Motorola		
			MC10563	† Motorola		
MC10593	Motorola					
Error Detection/Correction Circuit (TTL)	—	5	54F416	Fairchild	25	
			54F418	† Fairchild		
			54F630	Fairchild		
			54F631	Fairchild		
			74F416	Fairchild		
			74F418	Fairchild		
			74F630	Fairchild		
			74F631	Fairchild		
			9428	Fairchild		
			<b>MC74F2960</b>	<b>Motorola (822)</b>		30
			<b>SN74ALS790</b>	<b>Motorola (823)</b>		
			MM54ALS632	† National		35
			MM54ALS633	† National		
			MM54ALS634	† National		40
			MM54ALS635	† National		
			MM74ALS632	National		45
			MM74ALS633	National		
			MM74ALS634	National		50
			MM74ALS635	National		
			54F630	† Signetics (888)		55
			54F631	† Signetics (888)		
			74F630	Signetics		60
				(880,888)		
			74F631	Signetics		65
				(880,888)		
			SN54ALS632	† TI (1070,374)		70
			SN54ALS633	† TI (374,1070)		
SN54ALS634	† TI (374,1071)	75				
SN54ALS635	† TI (374,1071)					
SN54LS630	† TI (1069)	80				
SN54LS631	† TI (1069)					
SN54LS636	† TI (1072,1153)	85				
SN54LS637	† TI (1072,1153)					
SN74ALS632	TI (374,1070)	90				
SN74ALS633	TI (374,1070)					
SN74ALS634	TI (374,1071)	95				
SN74ALS635	TI (374,1071)					
SN74LS630	TI (1069)	100				
SN74LS631	TI (1069)					
SN74LS636	TI (1072,1153)	105				
SN74LS637	TI (1072,1153)					
MB1412A	Fujitsu	7				
Error Detection/Correction Circuit (TTL), Serial Burst	—	5	54F430	Fairchild	65	
			74F430	Fairchild		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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# IC MASTER

## INTERFACE—Error Checking Circuits (Cont'd)

Function	Max. Serial Data Rate, MHz	Supply Voltage, V	Device	Source	Line
<b>Error Checking Circuits</b>			<b>(Cont'd)</b>		
Error Detection/Correction Circuit, 32-Bit Memory	—	5	<b>54F418</b> <b>74F418</b>	† Fairchild Fairchild	(757) (757)
Error Pattern Register	3	5	MC8501	Motorola	
Expandable Error Checker and Corrector	—	5	<b>DP8400</b> DP8400	National TI	(1514)
LRCC Data Register	3	5	MC8502 <b>SSI8502</b>	Motorola SiliconSys	(4208)
Polynomial Generator	4 *	5	MC8506	Motorola	
Polynomial Generator/Checker	3.5 *	5	<b>MC2653</b> <b>MC68653</b>	Motorola Motorola	(1512) (1506, 1512)
			MC8503 <b>SCN68653</b>	Motorola Signetics	(1671)
	4	5	<b>SCN2653</b>	Signetics	(1671)
Universal Polynomial 4-Bit Generator	17 *	5	MC8504	Motorola	
Single Error Hamming Code Detector and Generator	—	5	MC4041	Motorola	15

† Military Temperature Range (–55° to 125°C)

\* Typical Value

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INTERFACE-Keyboard Encoder-Decoders

No. of Keys	No. of Output Bits	Code	Max. Clock Rate, KHz	Supply Voltage, V	Comments	Device	Source	Line
<b>Keyboard Encoder- Decoders</b>								
16	4	Binary	—	3-15	2 key rollover, 3 state output	MM54C922 MM74C922	† National National	5
		External ROM Programmable	—	5	Antibounce, mute, interlock	M190	SGS	
		1 of 16	—	5	Strobe output, key rollover output, 2 of 8 keyboard to binary encoder, one of four row inputs and column inputs (telephone key pads) give binary output, strobe.	HD0165 MC14419	Harris (2979) Motorola (3536)	
20	5	Binary	—	3-15	2 key rollover, 3 state output	MM54C923 MM74C923	† National National	
85	8	ASCII/HEX	1000	4-10.5	Scans and generates code for 53 key ASCII plus 32 non-ASCII keys	CDP1871A CDP1871AC	RCA RCA	(1595) (1595)
				4-6.5	Scans and generates code for 53 key ASCII plus 32 non-ASCII keys			
88	8 plus Parity	Mask, Programmable	100	— 12.5	Programmable parity, strobe width, strobe delay. Two key rollover. 8 x 11 matrix, 3 levels.	KR2376	SMC	10
90	10	External ROM Programmable	100	— 12.5	9 x 10 matrix, 4 Mode, 2 or N key rollover	KR3600-PRO	SMC	
		Mask, Programmable	100	— 12.5	9 x 10 matrix, 4 Mode, 2 or N key rollover	KR3600	SMC	
112	10	Mask Programmable	66	5	112 bits for internal programming of function keys	AY3-4592	Gi	
128	8	Mask Programmable	400	5	16 x 18 matrix, 8-Bit bus interface, 4 rollover modes, UART on chip	SCN2671A	Signetics (1678)	15
	9	Mask Programmable	100	5	16 x 18 matrix, three-state I/O, 2 or 3 key alarm	MSM3914A	OKI (4118)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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# IC MASTER

## INTERFACE-Line Circuits

No. Per Device	Output	Party Line	Supply Voltage, V	Comments	Device	Source	Line		
<b>Line Drivers—Single Ended</b>									
2	High Current	Yes	5	Coax/Twisted Pair	MC8T23	Motorola	5		
					DS75123	National			
	<b>N8T23</b>	<b>Signetics (876)</b>							
	N8T23	TI							
	SN75123	TI							
					Emitter Follower for Coax/Twisted Pair	MC8T13		Motorola	
						DS55121		† National	
						DS75121		National	
						<b>N8T13</b>		<b>Signetics (876)</b>	
						<b>S8T13</b>		<b>Signetics (876)</b>	
					N8T13	TI			
					SN55121	† TI			
					SN75121	TI			
					μA9639C	TI			
					9636AC	Fairchild			
					9636AM	† Fairchild			
					μA9636AC	TI			
					<b>N8T15</b>	<b>Signetics (876)</b>			
					75150	Fairchild			
					DS75150	National			
					SN55150	† TI			
					SN75150	TI			
3	High Current	No	5	IBM360	HD2904	Hitachi	25		
					9616	Fairchild			
	9616E	Fairchild							
	9616M	† Fairchild							
	± 6 V	No	± 12	RS232B/C, CCITT, MIL 188	DS7832	† National			
					DS8832	National			
					DS8832	TI			
					Three-State	<b>N8T09</b>		<b>Signetics (876)</b>	
						<b>S8T09</b>		<b>† Signetics (876)</b>	
					2-Input NAND, 80 Ma	96101C		Fairchild	
					96101M	† Fairchild			
				4-Input AND, NAND	DS7831	† National			
					DS8831	National			
					DS8831	TI			
				0.15 to 4 V	No	5	IBM360/370	<b>MC3481</b>	<b>Motorola (3064,3065)</b>
								<b>MC3485</b>	<b>Motorola (3064,3065)</b>
				3.11	No	5	IBM360/370	SN75126	TI
								SN75130	TI
				-7 to 12	Yes	5	RS422, Three-State	SN75174	TI
				-7 to ± 12	Yes	5	RS422, Three-State	SN75172	TI
				± 4 to 6 V	No	± 5	RS423, RS422 with mode control	DS3691	National
								± 5.5	RS423, RS422 with mode control
					Yes	± 5	RS423, Three-State	AM26LS29C	AMD
								AM26LS29	Signetics
						± 5.5	RS423, Three-State	AM26LS29M	† AMD
				± 6 to 9 V	No	± 9 to ± 15	RS232C, CCITT V.24	<b>XR1488</b>	<b>Exar (3384)</b>
								μA1488	Fairchild
							HD75188	Hitachi	
							<b>MC1488</b>	<b>Motorola (3064)</b>	
							DS1488	National	
							MC1488	Signetics	
							SG1488	Silicon6	
							MC1488	TI	
							SN55188	† TI	
							SN75188	TI	
				± 4 to 6 V	No	± 5	RS423, RS422 with mode control	AM26LS30C	AMD
									AM26LS30
						± 5.5	RS423, RS422 with mode control	AM26LS30M	† AMD
6	NTDS	No	5, -15	NTDS	MOA-268B	† Interdesign	60		
					MOF1305B	† Interdesign			
				5, -5	NTDS Hex Driver				

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE—Line Circuits (Cont'd)

No. Per Device	Output	Party Line	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Drivers—Single Ended</b>					<b>(Cont'd)</b>		
8	TTL	Yes	5	Three-State (also see index for 74S244, 74LS244, 67S304, 67LS304)	54LS241	† Fairchild	
					54LS541	† Fairchild	
					74LS241	Fairchild	5
					74LS244	Fairchild	
					74LS541	Fairchild	5
					HD74LS241	Hitachi	
					HD74LS244	Hitachi	10
					M74LS241	Mitel	
					M74LS244	Mitsubishi	10
					<b>SN54LS241</b>	† <b>MMI (804)</b>	
					<b>SN54LS244</b>	† <b>MMI (804)</b>	15
					<b>SN54LS245</b>	† <b>MMI (804)</b>	
					<b>SN54LS341</b>	† <b>MMI (804)</b>	15
					<b>SN54LS344</b>	† <b>MMI (804)</b>	
					<b>SN54S241</b>	† <b>MMI (804)</b>	15
					<b>SN54S244</b>	† <b>MMI (804)</b>	
					<b>SN74LS241</b>	<b>MMI (804)</b>	20
					<b>SN74LS244</b>	<b>MMI (804)</b>	
					<b>SN74LS245</b>	<b>MMI (804)</b>	20
					<b>SN74LS341</b>	<b>MMI (804)</b>	
					<b>SN74LS344</b>	<b>MMI (804)</b>	20
					<b>SN74S241</b>	<b>MMI (804)</b>	
					<b>SN74S244</b>	<b>MMI (804)</b>	25
					SN54LS241	† Motorola	
					SN54LS244	† Motorola	25
					SN54LS541	† Motorola	
					SN74LS241	Motorola	30
					SN74LS244	Motorola	
					SN74LS541	Motorola	30
					DM54S241	† National	
					DM74LS241	National	35
					DM74LS244	National	
					DM74S241	National	35
					MM54C941	† National	
					MM74C941	National	35
					74LS241	Signetics	
					74LS244	Signetics	40
					74LS541	Signetics	
					<b>SN54ALS241</b>	† <b>TI (999)</b>	40
					<b>SN54ALS244</b>	† <b>TI (1000)</b>	
					<b>SN54ALS541</b>	† <b>TI (1050)</b>	45
					<b>SN54AS241</b>	† <b>TI (999)</b>	
					SN54LS241	† TI	45
					SN54LS244	† TI	
					<b>SN54LS541</b>	† <b>TI (1050)</b>	45
					SN54S241	† TI	
					<b>SN74ALS241</b>	<b>TI (999)</b>	50
					<b>SN74ALS244</b>	<b>TI (1000)</b>	
					<b>SN74ALS541</b>	<b>TI (1050)</b>	50
					<b>SN74AS241</b>	<b>TI (999)</b>	
					SN74LS241	TI	55
					SN74LS244	TI	
					<b>SN74LS541</b>	<b>TI (1050)</b>	55
					SN74S241	TI	
				Three-State, Inverting	<b>54F240</b>	† <b>Fairchild (729)</b>	55
					54LS240	† Fairchild	
					54LS540	† Fairchild	60
					<b>74F240</b>	<b>Fairchild (729)</b>	
					74LS240	Fairchild	60
					74LS540	Fairchild	
					M74LS240	Mitsubishi	65
					<b>SN54LS240</b>	† <b>MMI (804)</b>	
					<b>SN54LS340</b>	† <b>MMI (804)</b>	65
					<b>SN54S240</b>	† <b>MMI (804)</b>	
					<b>SN74LS240</b>	<b>MMI (804)</b>	65
					<b>SN74S240</b>	<b>MMI (804)</b>	
					SN54LS240	† Motorola	70
					SN54LS540	† Motorola	
					SN74LS240	Motorola	70
					SN74LS540	Motorola	

† Military Temperature Range (−55° to 125°C)

\* Typical Value

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**IC MASTER**

**INTERFACE-Line Circuits (Cont'd)**

No. Per Device	Output	Party Line	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Drivers—Single Ended</b>					<b>(Cont'd)</b>		
8	TTL	Yes	5	Three-State, Inverting	DM54S240 DM74S240 54LS540 74LS240 74LS540 <b>SN54ALS240</b> † TI (998) <b>SN54ALS540</b> † TI (1050) <b>SN54AS240</b> † TI (998) SN54LS240 † TI <b>SN54LS540</b> † TI (1050) SN54S240 † TI <b>SN74ALS240</b> TI (998) <b>SN74ALS540</b> TI (1050) <b>SN74AS240</b> TI (998) SN74LS240 TI <b>SN74LS540</b> TI (1050) SN74S240 TI	National National Signetics Signetics Signetics TI TI TI TI TI TI TI TI TI TI	(Cont'd) 5 10 15 20 25
16	TTL	Yes	5	Three-State	<b>SN54LS365A</b> † TI (1023) <b>SN54LS367A</b> † TI (1024) <b>SN74LS365A</b> TI (1023) <b>SN74LS367A</b> TI (1024)	TI TI TI TI	20
				Three-State, Inverting	<b>SN54LS366A</b> † TI (1023) <b>SN54LS368A</b> † TI (1024) <b>SN74LS366A</b> TI (1023) <b>SN74LS368A</b> TI (1024)	TI TI TI TI	25
<b>Line Drivers—Differential</b>							
See also Drivers under digital logic families							
2	High Current CMOS	No	3-15	CMOS 50 mA, 4-Input AND, NAND	MM78C30 MM88C30	† National National	
	High Current TTL	No	5	40 mA, Active Pull-up/Pull-down 40 mA, Open Collector/Active Pull-up	9612C 9614C 9614M DS55114 DS75114 9614C SN55114 SN75114	Fairchild Fairchild † Fairchild † National National TI † TI TI	30 35
				40 mA, RS422	SN55128 SN75158	† TI TI	
				40 mA, 4-Input AND, NAND	54S140 74S140 HD74S140 DM74S140 DS7830 DS8830 54S140 74S140 DS8830 <b>SN54S140</b> † TI (969) SN55183 † TI <b>SN74S140</b> TI (969) SN75183 TI	† Fairchild Fairchild Hitachi National † National National † Signetics (531) Signetics TI † TI (969) † TI TI (969) TI	40 45 50
		Yes	5	Three-State	DS7831 DS8831 DS8831	† National National TI	
				DS7831 w/o Vcc Clamp	DS7832 DS8832	† National TI	55
				40 mA, Open Collector/Active Pull-Up, Three-State	DS55113 DS75113 SN55113 SN75113	† National National † TI TI	60
				40 mA, RS422, Three-State	9634C	Fairchild	

(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE-Line Circuits (Cont'd)

No. Per Device	Output	Party Line	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Drivers—Differential</b>					<b>(Cont'd)</b>		
2	High Current TTL	Yes	5	40 mA, RS422, Three-State	9634M DS8832 SN75159	† Fairchild National TI	(Cont'd)
	2.5 to 3.5	No	5	High Speed	μA9638C	TI	
	3.5 mA	Yes	± 5	Twisted Pair	HD75109 MC75109 SN55109A <b>SN74109A</b>	Hitachi Motorola † TI <b>TI</b>	5 (958)
	6.5 mA	Yes	± 5	Twisted Pair, Level Shifting	55110A 75110A HD75110 MC75110 MC75S110 SN55110A SN75110A	† Fairchild Fairchild Hitachi Motorola Motorola † TI TI	10 15
	18 mA	Yes	± 5	Higher Current 75110	SN75112	TI	
	± 3 V	Yes	± 5	RS422 at Low Data Rates, RS423	AM26LS30C AM26LS30	AMD Signetics	
			± 5.5	RS422 at Low Data Rates, RS423	AM26LS30M	† AMD	
	± 5V	No	± 15, ± 5	ARING 429, 1000K bits data rate	<b>HS3182</b>	† Harris	(4644) 20
4	High Current CMOS	No	3-15	CMOS, 25 mA	MM78C29 MM88C29	† National National	
	High Current TTL	No	5	50 Ohm Lines	74128 <b>SN74128</b>	Signetics TI	(861) (964)
				75 Ohm Lines	54128 <b>SN54128</b>	† Signetics † TI	(861) (964) 25
		Yes	5	RS-422, RS-423, Three-State	MC3487 DS3487 MC3487	Motorola National TI	(3064)
				Three-State	DS1688	† National	30
	11 mA	Yes	± 5	Quad 75110	MC3453A	Motorola	
	20 mA	Yes	5	MIL 188-114, Three-State	DS1692 DS3692	† National National	
				RS-422, Three-State	AM26LS31C AM26LS31M	AMD † AMD	
					AM26LS31	Motorola	(3064)
					AM26LS31C	Motorola	(3064)
					AM26LS31M	† Motorola	(3064)
					DS26LS31C	† National	
					DS26LS31M	National	
					AM26LS31	Signetics	
					AM26LS31C	TI	
					AM26LS31M	† TI	
	40 mA	Yes	5	RS-422, Fed. 1020, Three-State	SN75151 SN75153	TI TI	45

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## INTERFACE—Line Circuits (Cont'd)

No. Per Device	Receiver Input Threshold	Common Mode Voltage, V	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Receivers—Single Ended</b>							
Also see Line Receivers-Differential							
2	±0.1 V (ref)	5	5	Ext. Ref. Adj. 1.5 to 3.5 V with Optional Internal 2.5 V Reference	SN55142A	† TI	5
				SN55143A	† TI		
				Ext. Reference Adjustable from 1.5 to 3.5 V	SN75142A	TI	
					SN75143A	TI	
					SN55140	† TI	
					SN55141	† TI	
					SN75140	TI	
					SN75141	TI	
	±2.0	5		Hysteresis, EIA/MIL RS232B	<b>N8T16</b>	<b>Signetics (876)</b>	10
					DS7822	† National	
					DS8822	National	
3	0.7 to 1.7	5	5	Hysteresis, IBM360/370	MC8T24	Motorola	15
					DS75124	National	
					<b>N8T24</b>	<b>Signetics (876)</b>	
					SN75124	TI	
	0.75 to 2.25	5		Hysteresis, RS232C, CCITT V.24	9617C	Fairchild	
	0.8 to 2.0 V	5		Hysteresis, High-Speed	MC8T14	Motorola	20
					DS55122	† National	
					DS75122	National	
					N8T14	Signetics	
					SN55122	† TI	
					SN75122	TI	
	0.86 to 2.40	5		High-Speed	HD2915	Hitachi	
	1.15 to 1.55	5		IBM360	HD2905	Hitachi	25
4	0.75 to 1.5 V	5	5	RS232C, Programmable Threshold, Hysteresis	μA1489	Fairchild	30
					HD75154	Hitachi	
					<b>MC1489</b>	<b>Motorola (3064)</b>	
					DS1489	National	
					MC1489	Signetics	
					SG1489	SiliconG	
					SN55189	† TI	
					SN75189	TI	
	0.75 to 2.25	5		RS232C, Programmable Threshold, Wider Hysteresis than 1489	<b>XR1489A</b>	<b>Exar (3384)</b>	35
					μA1489A	Fairchild	
					HD75189	Hitachi	
					<b>MC1489A</b>	<b>Motorola (3064)</b>	
					DS1489A	National	
					MC1489A	Signetics	
					SG1489A	SiliconG	
					SN55189A	† TI	
	0.97 to 2.65	5		Hysteresis, 120 Ohm System	DS7836	† National	
	1.05 to 2.5	5		Hysteresis, 120 Ohm System	DS8836	National	
	1.2 to 1.8	5		120 Ohm System, No Hysteresis, NOR Input	DS7640	† National	
	1.3 to 1.7	5		120 Ohm System, No Hysteresis, NOR Input	DS8640	National	45
	1.7	5		No Hysteresis, NOR Input	96106	Fairchild	
	1.75 to 2.25 V	5		RS232C, Programmable Threshold, Hysteresis	SN75189A	TI	
	5 to 7.5	15		Hysteresis, Interface to CMOS	367A	TeledyneS	50
					367M	† TeledyneS	
				Hysteresis, Open Collector, Interface to CMOS	368A	TeledyneS	
	5.5 to 8	12		Hysteresis, Open Collector, Interface to CMOS	368C	TeledyneS	
				Hysteresis, Interface to CMOS	367B	TeledyneS	
					367C	† TeledyneS	
	±3/0.8 to 3	5 or 12		RS232C, Hysteresis, Fail Safe Option	55154	† Fairchild	55
					75154	Fairchild	
					DS75154	National	
					SG75154	SiliconG	
					<b>SN54154</b>	<b>† TI (973)</b>	
					SN75154	TI	
6	0.97 to 2.65	5	5	Hysteresis, 120 Ohm System	DS7837	† National	60
					MC3437	Motorola	
	1.05 to 2.50			Hysteresis, 120 Ohm System	DS8837	National	
					<b>N8T37</b>	<b>Signetics (876)</b>	
	-1 to 4.5/-1.5 to 13	5		NTDS	MOB-272	† Interdesign	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Line Circuits (Cont'd)

No. Per Device	Receiver Input Threshold	Common Mode Voltage, V	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Receivers—Single Ended</b>					<b>(Cont'd)</b>		
7	0.7 to 1.7		5	IBM360/370	MC75125 MC75127 DS75125 DS75127 SN75125 SN75127	Motorola (3064) Motorola National National TI TI	5
8	0.7 to 1.7		5	IBM360/370	MC75128 MC75129 DS75128 DS75129 SN75128 SN75129	Motorola Motorola National National TI TI	10
<b>Line Receivers—Differential</b>							
See also Receivers—listed under Digital-ECL and Digital-HNIL/HTL Miscellaneous sections							
2	± 0.010	± 3	± 5	10 mV, MOS Sense, Active Pull-up	DS75207 DS75208 SN75207	National National TI	15
				10 mV, MOS Sense, Open Collector	SN75206	TI	
				10 mV, MOS Sense, Three-State	DS3604	National	
				75207 with Diode Protected Input Stage.	SN75207B	TI	
				75208 with Diode Protected Input Stage	SN75208B	TI	20
	± 0.025	± 3	± 5	25 mV, Active Pull-up	55107A 75107A HD75107A MC55107 MC75107	† Fairchild Fairchild Hitachi † Motorola Motorola	25
				25 mV, Active Pull-Up	DS55107 DS75107	† National National	
				25 mV, Active Pull-up	SN55107A SN75107A	† TI TI	
				25 mV, Open Collector	HD75108A MC55108 MC75108 DS55108 DS75108 SN55108A SN75108A	Hitachi † Motorola Motorola † National National † TI TI	30
				25 mV, Three-State 55107	DS1603 DS3603	† National National	35
				55107A with Diode Protected Input Stage	75107B DS55107 DS75107 SN55107B SN75107B	Fairchild † National National † TI TI	40
				55108A with Diode Protected Input Stage	75108B DS55108 DS75108 SN55108B SN75108B	Fairchild † National National † TI TI	45
	± 0.2/0.3	± 10/15	5	Twisted Pair, ± 15 V CMV, Response Control	DS78LS120 DS88LS120	† National National	50
		± 15	5	CMOS Compatible, Response Control	DS78C120 DS88C120	† National National	
	± 0.2/0.5	± 7/15	5	RS232, RS422/3	9637AC 9637AM SN55157 SN75157 μA9637AC	Fairchild † Fairchild † TI TI TI	55
	± 0.3/0.3	± 0/15	5	± 15 V CMV, Response Control	DS78L20	† National	
	± 0.5/1	± 0/15	5	± 15 V CMV, Response Control	9615C 9615M DS55115 DS75115 9615C	Fairchild † Fairchild † National National TI	60

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† Military Temperature Range (−55° to 125°C)

\* Typical Value

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**INTERFACE-Line Circuits (Cont'd)**

No. Per Device	Receiver Input Threshold	Common Mode Voltage, V	Supply Voltage, V	Comments	Device	Source	Line	
<b>Line Receivers—Differential</b>					<b>(Cont'd)</b>			
2	± 0.5/1	± 0/15	5	± 15 V CMV, Response Control	SN55115 SN75115	† TI TI	5	
		± 3/15	5	Twisted Pair, ± 3/15 V CMV, Response Control	DS7820A DS8820A DS8820 SN55152 SN55182 SN75182	† National National TI † TI † TI TI		
	± 10/15	5	Twisted Pair, ± 15 V CMV, Response Control	DS7820 DS78C20 DS8820 DS88C20	† National † National National National	10		
4	± 0.5/3	± 25	± 12	Adjusts RS232C/MIL-188, ± 25 V CMV, Hysteresis	SN75152	TI	15	
			± 15/12-24	Quad Comparator, 3mV Hysteresis	ESM1600 ESM1602	Thomson-CSF Thomson-CSF		
	± 0.025	± 3	± 5	Four 75107, Active Pull-up	MC3450 DS1650 DS3650	Motorola † National National		
				Four 75108, Open Collector	MC3452 DS1652 DS3652	Motorola † National National		20
	± 0.2	± 3	± 5	Three-State, RS422/423	MC3486 DS3486 MC3486	Motorola National TI		25
				Three-State, RS-422/423	AM26LS32C AM26LS32M DS26LS32C DS26LS32M AM26LS32C AM26LS32M	AMD † AMD National † National TI † TI		
	± 12	± 5	± 5	Three-State, RS-422	SN55173 SN75173 SN75175	† TI TI TI		30
				50 mV Hysteresis	DS1689 DS1690 DS3689 DS3690	† National † National National National		
	± 0.5	± 15	± 5	Three-State	AM26LS33C AM26LS33M DS26LS33C DS26LS33M AM26LS33C AM26LS33M	AMD † AMD National † National TI † TI		40
				NTDS	MOF1623B	† Interdesign		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE-Line Circuits (Cont'd)

No. Per Device	Receiver Input Threshold, V	Output	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Transceivers—Single Ended</b>							
4	0.05 to 2.50	TTL	5	Open Collector, 1 V Hysteresis	MC3438 DS8838 <b>IN8738</b>	Motorola National <b>Signetics (878)</b>	5
	0.21 to 1.84	TTL	5	Open Collector (Inverting), Common Enable	96103M	† Fairchild	
	0.4 to 2.05	TTL	5	Open Collector, Hysteresis	AM26S12AC AM26S12AM	AMD † AMD	
	0.5 to 2.0	TTL	5	Bus Transceiver, Tridirectional, Open Collector	<b>SN54LS440</b> <b>SN54LS441</b> <b>SN54LS448</b>	† TI (1035,1036) † TI (1035,1036) † TI (1039)	10
				Bus Transceiver, Tridirectional, Three-State	<b>SN54LS442</b> <b>SN54LS443</b> <b>SN54LS444</b>	† TI (1035,1036) † TI (1035,1036) † TI (1035,1036)	
	0.6 to 2.0	TTL	5	Bus Transceiver, Tridirectional, Open Collector	<b>SN74LS440</b> <b>SN74LS441</b> <b>SN74LS448</b>	TI (1035,1036) TI (1035,1036) TI (1039)	15
				Bus Transceiver, Tridirectional, Three-State	<b>SN74LS442</b> <b>SN74LS443</b> <b>SN74LS444</b>	TI (1035,1036) TI (1035,1036) TI (1035,1036)	
	0.7 to 2.0	TTL	5	General Purpose Interface Bus, Open Collector, for MOS Input	MC3446A MC3446	Motorola (3064) TI	20
				Three-State, Bus Transceiver	AM2915AM AM2916AM AM2917AM	† AMD (1411) † AMD (1411) † AMD (1412)	
	0.8 to 1.8	TTL/MOS	5	Three-State, Bus Transceiver	54LS242 54LS243	† Fairchild † Fairchild	25
				Three-State, Bus Transceiver, Parity Generator/Checker	SN54LS242 SN54LS243 SN54LS243 SN54LS242 SN54LS242 SN54LS243	† Motorola † Motorola † Motorola † TI (999) † TI (1000) † TI	
	0.8 to 2.0	TTL	5	Three-State, Hysteresis	SN54LS242 SN54LS243	† TI † TI	30
				Bus Transceiver, Individual Direction Controls	<b>SN54LS446</b> <b>SN54LS449</b>	† TI (1038) † TI (1039)	
	0.8 to 2.0	TTL	5	General Purpose Interface Bus, Bidirectional Bus Transceiver, Three-State	AM3448A μA3448A <b>MC3448A</b>	AMD Fairchild <b>Motorola (3064)</b>	35
				Bus Transceiver, Individual Direction Controls	<b>SN74LS446</b> <b>SN74LS449</b>	TI (1038) TI (1039)	
	0.8 to 2.0	TTL	5	General Purpose Interface Bus, Open Collector, 100 mA Output	DS8641 MC3440A MC3441A MC3443A DS3662 DS7641 DS8641 MC3443	Motorola <b>Motorola (3064)</b> Motorola Motorola National † National National TI	40
				Three-State, Bus Transceiver	AM2915AC AM2916AC AM2917AC	AMD (1411) AMD (1411) AMD (1412)	
	0.8 to 2.0	TTL	5	Three-State, Bus Transceiver, Parity Generator/Checker	<b>SN74S226</b>	TI (996)	50
				Three-State, Dual Rank Latches			
	0.8 to 2.0	TTL	5	Three-State, Hysteresis	74LS242 74LS243 HD74LS242 M74LS242 M74LS243 SN74LS242 SN74LS243 DM74LS242 DM74LS243 74LS242 74LS243 <b>SN74ALS242</b> <b>SN74ALS243</b> SN74LS242 SN74LS243	Fairchild Fairchild Hitachi Mitel Mitel Motorola Motorola National National Signetics Signetics TI (999) TI (1000) TI TI	60
				Three-State (Inverting)	μA8T26A	Fairchild	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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**IC MASTER**

**INTERFACE-Line Circuits (Cont'd)**

No. Per Device	Receiver Input Threshold, V	Output	Supply Voltage, V	Comments	Device	Source	Line			
<b>Line Transceivers—Single Ended</b>					<b>(Cont'd)</b>					
4	0.8 to 2.0	TTL	5	Three-State (Inverting)	$\mu$ A8T26AM	† Fairchild	5			
					HD268T26	Hitachi				
					MC6880	Motorola				
					MC8T26A	Motorola				
					DS8T26A	National				
					DS8T26AM	† National				
					<b>N8T126</b>	<b>Signetics (876)</b>				
					<b>N8T127</b>	<b>Signetics (876)</b>				
					<b>N8T26A</b>	<b>Signetics (876)</b>				
					<b>S8T126</b>	<b>† Signetics (876)</b>				
					<b>S8T127</b>	<b>† Signetics (876)</b>				
						SN75136		TI		
4	0.8 to 2.0	TTL	5	Three-State (Non-Inverting)	$\mu$ A8T28	Fairchild	15			
					$\mu$ A8T28M	† Fairchild				
					MC6889	Motorola				
					MC8T28	Motorola				
					DS8T28	National				
					DS8T28M	† National				
					<b>N8T128</b>	<b>Signetics (876)</b>				
					<b>N8T129</b>	<b>Signetics (876)</b>				
					<b>N8T28</b>	<b>Signetics (876)</b>				
					<b>S8T128</b>	<b>† Signetics (876)</b>				
					<b>S8T129</b>	<b>† Signetics (876)</b>				
				4	0.97 to 2.65	TTL		5	Open Collector, 1 V Hysteresis	AM7838
	DS7838	† National								
1.05 to 2.50	TTL	5	Inverting 7833/8833				DS7835		† National	
							DS8835		National	
			Inverting 7839/8839				DS7834		† National	
							DS8834		National	
							<b>N8T34</b>		<b>Signetics (876)</b>	
			Three-State, NOR Gate, Transmit Disable, Hysteresis				DS7839		† National	
							DS8839		National	
			Three-State, T/R Disables, Hysteresis				DS7833		† National	
							DS8833		National	
4	1.5 to 2.4	TTL	5				Open Collector, 100 mA Output, Parity Generator/Checker		<b>AM2907M</b>	<b>† AMD (1406)</b>
				Q-bus Compatible 2907	<b>AM2908M</b>	<b>† AMD (1406)</b>				
				2-Input, Open Collector 100 mA Drivers	<b>AM2905M</b>	<b>† AMD (1405)</b>				
				2-Input, Open Collector 100 mA Drivers, Parity Generator/Checker	<b>AM2906M</b>	<b>† AMD (1405)</b>				
4	1.5 to 3.2	TTL	5	Open Collector, 100 mA Output	SN55138	† TI	40			
				1.6 to 1.8	TTL	5		Open Collector, Hysteresis	AM26S12C	AMD
					AM26S12M	† AMD				
4	1.6 to 2.3	TTL	5	Open Collector, 100 mA Output, Parity Generator/Checker	<b>AM2907C</b>	<b>AMD (1406)</b>	45			
				Q-bus Compatible 2907	<b>AM2908C</b>	<b>AMD (1406)</b>				
				2-Input, Open Collector 100 mA Drivers	<b>AM2905C</b>	<b>AMD (1405)</b>				
				2-Input, Open Collector 100 mA Drivers, Parity	<b>AM2906C</b>	<b>AMD (1405)</b>				
4	1.6 to 2.4	TTL	5	Open Collector, 100 mA Output	AM26S10M	† AMD	50			
					AM26S11M	† AMD				
					9640M	† Fairchild				
					DS26S10M	† National				
					DS26S11M	† National				
					AM26S10M	† TI				
	AM26S11M	† TI								
4	1.75 to 2.25	TTL	5	Open Collector, 100 mA Output	AM26S10C	AMD	55			
					AM26S11C	AMD				
					9640C	Fairchild				
					MC26S10	Motorola				
					MC26S11	Motorola				
					DS26S10C	National				
					DS26S11C	National				
					AM26S10C	TI				
					AM26S11C	TI				
						SN75138		TI		
4	1.8 to 2.9	TTL	5	Open Collector, 100 mA Output	SN75138	TI	60			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE-Line Circuits (Cont'd)

No. Per Device	Receiver Input Threshold, V	Output	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Transceivers—Single Ended</b>					<b>(Cont'd)</b>		
8	—	TTL/CMOS	5	General Purpose Interface Bus	DS3666 DS75160A DS75161A DS75162A	National National National National	
	TTL	TTL	5	Bidirectional Bus Transceiver, Three-State	DP7304B DP8304B DP8304	† AMD AMD National	5
	0.5 to 2.0	TTL	5	Bidirectional Bus Transceiver, Open Collector	54LS641 54LS642 74LS641 74LS641-1 74LS642 74LS642-1 SN54LS621 SN54LS622 SN54LS641 SN54LS642 SN54LS644	† Signetics † Signetics Signetics Signetics Signetics Signetics † TI (1066) † TI (1066) † TI (1074) † TI (1074) † TI (1074)	10 15
				Bidirectional Bus Transceiver, Three-State	DP7303 DP7307 DP7308 DP8303 DP8307 DP8308 SN54LS645 SN54LS645-1	† AMD † AMD † AMD AMD AMD AMD † MMI (804) † MMI (804)	20 25
					DP7303 DP7304B DP7307 DP7308 DP8303 DP8304B DP8307 DP8308 SN54ALS1645 SN54ALS245 SN54ALS645 SN54LS620 SN54LS623 SN54LS640 SN54LS643 SN54LS645 SN54SL245	† National † National † National † National National National National National † TI (1124) † TI (1001) † TI (1074) † TI (1066) † TI (1066) † TI (1074) † TI (1074,1074,1075) † TI (1074) † TI	30 35 40
	0.6 to 2.0	TTL	5	Bidirectional Bus Transceiver, Open Collector	SN74LS621 SN74LS622 SN74LS641 SN74LS641-1 SN74LS642 SN74LS642-1 SN74LS644 SN74LS644-1	TI (1066) TI (1066) TI (1074) TI (1074) TI (1074) TI (1074) TI (1074) TI (1074)	45 50
				Bidirectional Bus Transceiver, Three-State	SN74LS645 SN74LS645-1 74LS640 74LS640-1 74LS645 SN74LS245 SN74LS620 SN74LS623 SN74LS640 SN74LS640-1 SN74LS643 SN74LS643-1 SN74LS645 SN74LS645-1	MMI (804) MMI (804) Signetics Signetics Signetics TI TI (1066) TI (1066) TI (1074) TI (1074) TI (1074) TI (1074) TI (1074) TI (1074) TI (1074)	55 60
	0.7 to 2.0	TTL	5	Bidirectional Bus Transceiver, Open Collector	SN54LS641 SN54LS642	† Motorola † Motorola	65

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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**INTERFACE-Line Circuits (Cont'd)**

No. Per Device	Receiver Input Threshold, V	Output	Supply Voltage, V	Comments	Device	Source	Line
<b>Line Transceivers—Single Ended</b>					<b>(Cont'd)</b>		
8	0.7 to 2.0	TTL	5	Bidirectional Bus Transceiver/Register, Open Collector	SN54LS647 † TI	(1076)	5
					SN54LS649 † TI	(1076)	
				Bidirectional Bus Transceiver, Three-State	54LS245 † Fairchild		10
					SN54LS245 † Motorola		
					SN54LS640 † Motorola		
					SN54LS645 † Motorola		
					SN54LS646 † TI (1076)		
					SN54LS648 † TI (1076)		
8	0.8 to 2.0	TTL	5	Bidirectional Bus Transceiver, Open Collector	SN74LS641 Motorola		15
					SN74LS642 Motorola		
				Bidirectional Bus Transceiver/Register, Open Collector	SN74LS647 TI (1076)		20
					SN74LS649 TI (1076)		
				Bidirectional Bus Transceiver, Three-State	74LS245 Fairchild		25
					G74SC245 GTEMicro		
					HD74LS243 Hitachi		
					HD74LS245 Hitachi		
					M74LS245 Mitsubishi		
					SN74LS245 Motorola		
					SN74LS640 Motorola		
					SN74LS645 Motorola		
					DM74LS245 National		
					DS3667 † National		
					74LS245 Signetics		
					NS7125 Signetics (876)		
					NS7125 † Signetics (876)		
					SN54AS245 † TI (1001)		
					SN74ALS1645 TI (1124)		
					SN74ALS245 TI (1001)		
					SN74ALS645 TI (1074)		
					SN74AS245 TI (1001)		
					SN74LS646 TI (1076)		
					SN74LS648 TI (1076)		
				Bidirectional Bus Transceiver, Three-State, CMOS	MD54SC245 † Mitel		30
					MD74SC245 Mitel		
				General Purpose Interface Bus	MC3447 Motorola (3064,3066)		35
					SN75160A TI		
					SN75161A TI		
					SN75162A TI		
					SN75163 TI		
<b>Line Transceivers—Differential</b>							
1		TTL	5	Designed to meet proposed EIA Standard RS485	DS3695 National		40
					DS3696 National		
					DS3697 National		
					DS3698 National		
	0.8 to 2	TTL	5	Bidirectional Bus Transceiver, Three State	SN75176 Motorola (3064)		45
					SN75177 Motorola		
					SN75178 Motorola		
					SN75176A TI		
					SN75177 TI		
					SN75178 TI		
	± 0.5/± 1 V	TTL	5	Independent Three-State 55113 Driver and 55115 Receiver	SN55116 † TI		50
					SN75116 TI		
				Same as 55116 with Three-State Receiver	SN55118 † TI		55
					SN75118 TI		
				Same as 55117 with Three-State Receiver	SN55119 † TI		
					SN75119 TI		
				Three-State 8 Pin, 40 MA	SN55117 † TI		55
					SN75117 TI		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE—Memory and Peripheral Drivers

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Memory &amp; Peripheral Drivers</b>				Driver, Control Circuit for Fast Switching Transistors UAA4002 Thomson-CSF				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver) SN75469 TI ULN2004A TI			
Addressable Peripheral Driver (latched, 8-output driver) NE590 <b>Signetics (3625)</b> NE591 <b>Signetics (3625)</b> SE590 † Signetics				Driver, Half Bridge, 2 A SG1635 † SiliconG SG3635 SiliconG				Driver, 7 Channel, CMOS/TTL Input (hammer, lamp, relay driver) XR2003M † Exar (3383) XR2203 Exar 9667 Fairchild 9667M † Fairchild MC1413 Motorola (3064) PBD352303 RIFA L203 SGS ULN2003 Signetics SG2003 SiliconG SG2005 SiliconG SG2013 SiliconG SG2015 SiliconG SG2023 SiliconG SG2025 SiliconG SG3853 SiliconG ULN-2003A Sprague ULN-2005A Sprague ULN-2013A Sprague ULN-2015A Sprague ULN-2023A Sprague ULN-2025A Sprague ULS-2003H † Sprague ULS-2005H † Sprague ULS-2013H † Sprague ULS-2015H † Sprague ULS-2023H † Sprague ULS-2025H † Sprague SN75468 TI ULN2003A TI			
Bidirectional Interface Circuit, High Voltage TEA1512 Thomson-CSF				Driver, Open Collector/Emitter, for 150 mA (load connected to negative supply) PBD3520 RIFA				Driver, 7 Channel, MOS/TTL Input (hammer, lamp, relay driver) XR2201 Exar XR2201M † Exar 9665 Fairchild MC1411 Motorola (3064) PBD352301 RIFA L201 SGS SG2001 SiliconG SG2011 SiliconG SG2021 SiliconG SG3851 SiliconG ULN-2001A Sprague ULN-2011A Sprague ULN-2021A Sprague ULS-2001H † Sprague ULS-2011H † Sprague ULS-2012H † Sprague ULS-2021H † Sprague SN75466 TI ULN2001A TI			
BIMOS Latched Driver UCN-5810A Sprague UCN-5812A Sprague UCN-5818A Sprague UCS-4401H † Sprague UCS-4801H Sprague UCS-4810H Sprague				Driver, Serial Input/16-Bit Parallel Output, High-Voltage, High-Current Outputs TSC9403 TeledyneS (3101) TSC9404 TeledyneS (3101)				Driver, 7 Channel, PMOS Input (hammer, lamp, relay driver) XR2202 Exar XR2202M † Exar 9666 Fairchild 9666M † Fairchild MC1412 Motorola (3064) PBD352302 RIFA L202 SGS SG2002 SiliconG SG2012 SiliconG SG2022 SiliconG SG3852 SiliconG ULN-2002A Sprague ULN-2012A Sprague ULN-2022A Sprague ULS-2002H † Sprague ULS-2022H † Sprague SN75467 TI ULN2002A TI			
BIMOS Segment Latch Driver CA3208 RCA (3599)				Driver, Solenoid and Motor, to 40 V, 4 A SG3640 SiliconG SG3641 SiliconG				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver) (Cont'd)			
BIMOS Sequencer Driver CA3207 RCA (3599)				Driver, Switching Transistor LAS8100 Lambda LAS8101 Lambda				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bridged Motor Driver MD346 AnalogSys				Driver, to 80 V, 0.2 A DI445 Dionics				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Coil Driver DS3616 National				Driver, Single, 125 mA, for Relays, Motors, Lamps PBD3510 RIFA PBD3511 RIFA				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Coil Predriver SC42468 Motorola				Driver, Dual, to 80 V, 0.2 A DI446 Dionics				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Controller 7220 Intel SC42584 Motorola SC42585 Motorola				Driver, Dual, 2-Input, Sink or Source 500 mA SG1627 † SiliconG SG3627 SiliconG				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Controller (chevron) SN54LS361 † TI				Driver, 1 Channel (lamp relay driver) TDE3207 Thomson-CSF				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Controller (t-bar) SN54LS360 † TI				Driver, 2 Channel (hammer), to 35 V, 1 A SG3700 SiliconG				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Current Pulse Generator 7230 Intel				Driver, 2 Channel (solenoid and motor driver, 4 A) SG3638 SiliconG SG3638A SiliconG SG3639 SiliconG SG3639A SiliconG SG3643 SiliconG SG3643A SiliconG				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Function Driver DS3615 National DS3618 National				Driver, 4 Channel, 40 V, 2 A SG3637 SiliconG				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Predriver 7250 Intel SI7250 Siliconix				Driver, 5 Channel, CMOS/PMOS Input (lamp, relay driver, load to negative supply) UDN-2956A Sprague				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Sense Amplifier MC34044 Motorola DS3617 National				Driver, 5 Channel, CMOS/TTL Input (lamp, relay driver, load to negative supply) UDN-2957A Sprague				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory (Single) Operation Drivers MC34046S Motorola MC34047S Motorola				Driver, 5 Channel Darlington, to 400 mA XR2200 Exar XR2200M † Exar LB1287 Sanyo LB1288 Sanyo				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Dual Formatter/Sense Amplifier 7424 Intel				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver) MD402 AnalogSys XR2204 Exar XR2204M † Exar 9668 Fairchild 9668M † Fairchild MC1416 Motorola (3064) PBD352304 RIFA ULN2004 Signetics SG2004 SiliconG SG2014 SiliconG SG2024 SiliconG ULN-2004A Sprague ULN-2014A Sprague ULN-2024A Sprague ULS-2004H † Sprague ULS-2014H † Sprague ULS-2024H † Sprague				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Bubble Memory Quad VMOS Drive Transistors 7254 Intel				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Clock Generator/Oscillator, to 10 MHz, 8 and 1 Divider, for Microprocessors ICM7209 † Intersil (3015)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Data Acquisition Controller (intelligent) for A/D Converter CY600 Cybernetic				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
DC Motor Switch Mode Control UAA4004 Thomson-CSF				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Disc Memory Read/Write μPC751 NEC (3543) μPC752 NEC (3543)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Disk Drive, Disk Data Controller DP8466 National				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Disk Drive, Disk Pulse Detector DP8464 National (1514)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Disk Drive, Encoder/Decoder, 2, 7 DP8463 National				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			
Disk Drive, MFM Data Separator DP8460 National				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)				Driver, 7 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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**INTERFACE-Memory and Peripheral Drivers (Cont'd)**

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line		
<b>Memory &amp; Peripheral Drivers (Cont'd)</b>													
Driver, 8 Channel, CMOS/PMOS Input (hammer, lamp, relay driver)	ULN2804	Motorola	5	Gated Decoder, for SSI104/5 SSI106	SiliconSys		60	MOS Dynamic Memory, 16 K Address Multiplexer and Refresh Counter	96LS42C	Fairchild	105		
	ULN-2804A	Sprague			Hall Effect Digital Latches (logic to electromechanical interface)	UGN-3075		Sprague				96LS42M	† Fairchild
	ULN-2814A	Sprague				UGN-3076		Sprague				MC3242A	Motorola
	ULN-2824A	Sprague				UGS-3075		† Sprague		MOS Dynamic RAM Controller	8202A	Intel	110
	ULS-2804H	† Sprague			Hammer Driver (to 6 A-pulsed output)	DH0028C		National			8203	Intel	
	ULS-2814H	† Sprague									8205	Intel	
	ULS-2824H	† Sprague		High Current Switch Driver (to drive high power, high speed NPN switching transistors)	S61629	† SiliconG			<b>TMS4500A</b>	<b>TI (4221)</b>			
Driver, 8 Channel, CMOS/PMOS Input (lamp, relay driver)	UDN-2982A	Sprague	10		S63629	SiliconG			<b>VL4500A</b>	<b>VTI (4254)</b>	110		
	UDN-2984A	Sprague			High-Voltage, High-Current Source Driver	UDQ-2956R	Sprague			<b>WD8207</b>		<b>Western (4262)</b>	
	UDS-2982H	† Sprague				UDQ-2957	Sprague		MOS Dynamic RAM Controller/Driver	DP8408	MMI	115	
	UDS-2984H	† Sprague		High-Voltage Source Drivers	UDN-6510A	Sprague			<b>SN74S408</b>	<b>MMI (819)</b>			
Driver, 8 Channel, CMOS/TTL Input (hammer, lamp, relay driver)	ULN2803	Motorola	15		UDN-6514A	Sprague			<b>SN74S408-3</b>	<b>MMI (819)</b>			
	<b>NE5090</b>	<b>Signetics (3624)</b>			Memory Driver, FIFO RAM Controller	<b>MBX60</b>	<b>Signetics (1647)</b>			DP8408	National		
	SE5090	Signetics				<b>S8X60</b>	† <b>Signetics (527,1647)</b>		MOS Dynamic RAM Controller/Driver, Multimode	74S409	MMI	120	
	ULN-2803A	Sprague			Memory Driver Dual, 400 mA Sink/Source, Decode (for magnetic memories)	DS3629	National			DP8409	National		
	ULN-2805A	Sprague							MOS Dynamic RAM Controller/Driver, Two Port	<b>74LS764</b>	<b>Signetics (895)</b>		
	ULN-2813A	Sprague			Memory Driver, Dual 600 mA Sink/Source	MC55325	† Motorola			CY500	Cybernetic	70	
	ULN-2815A	Sprague				MC75325	Motorola		Motor Controller (intelligent) for 4-Phase Stepper Motors	CY512	-Cybernetic		
	ULN-2823A	Sprague				DS55325	† National		Multi-Mode Dynamic RAM Controller/Driver	<b>DP8408</b>	<b>National (1514)</b>	75	
	ULN-2825A	Sprague				DS75325	National			VF01	Supertex		
	ULS-2803H	† Sprague				S655325	† SiliconG		Power Peripheral Driver, CMOS/TTL Input (to 150 V, versions: 0.01 to 16 A)	VF02	Supertex		
	ULS-2805H	† Sprague			S675325	SiliconG			VF03	Supertex	125		
	ULS-2813H	† Sprague			SN55325	† TI			VF12	Supertex			
	ULS-2815H	† Sprague			SN75325	TI			VF13	Supertex			
	ULS-2823H	† Sprague						Printer Controller, for 5x7 Dot Matrix Printers	CY480	Cybernetic	80		
	ULS-2825H	† Sprague		Memory Driver, Quad, 600 mA Sink	S655326	† SiliconG			HD2919	Hitachi			
Driver, 8 Channel, CMOS/TTL Input (lamps, relay driver)	UDN-2981A	Sprague	30		S655327	† SiliconG		Printer Driver, 5 Channel			130		
	UDN-2983A	Sprague				S675326	SiliconG		Printer Solenoid Driver	DS3654		National	
	UDS-2981H	† Sprague				SN55326	† TI		Printing Calculator Circuits	DS8654	National	135	
	UDS-2983H	† Sprague			SN55327	† TI			DS8656	National			
Driver, 8 Channel, MOS/TTL Input (hammer, lamp, relay driver)	ULN2801	Motorola	35		SN55328	† TI			DS8692	National			
	ULN-2801A	Sprague				SN55329	† TI			DS8693	National		
	ULN-2811A	Sprague				SN75326	TI			DS8694	National		
	ULN-2821A	Sprague			Memory Timing Controller (with EDAC)	AM2969	AMD		Relay and Lamp Driver, Dual, to 32 V, 2/4 A	TDE1777	Thomson-CSF		
	ULS-2801H	† Sprague				<b>MC74F2969</b>	<b>Motorola (822)</b>		Relay and Lamp Driver, to 30 V, 0.15 A	TDE1647	Thomson-CSF		
	ULS-2811H	† Sprague							Relay and Lamp Driver, to 30 V, 0.3 A	TDE1607	Thomson-CSF		
	ULS-2821H	† Sprague						Relay and Lamp Driver, to 45 V, 0.3 A	TDE1647A	Thomson-CSF	140		
Driver, 8 Channel, PMOS Input (hammer, lamp, relay driver)	ULN2802	Motorola	40					Relay and Lamp Driver, to 50 V, 0.3 A	TDE1737	Thomson-CSF			
	ULN-2802A	Sprague			Memory Timing Controller (without EDAC)	AM2970	AMD		Relay and Lamp Driver, to 50 V, 0.5 A	TDE1767	Thomson-CSF		
	ULN-2812A	Sprague				<b>MC74F2970</b>	<b>Motorola (822)</b>		Relay and Lamp Driver, to 60 V, 0.3 A	TDE1787A	Thomson-CSF		
	ULN-2822A	Sprague			Modified Frequency Modulation Decoder Data Separator	<b>DP8460</b>	<b>National (1514)</b>		Relay and Lamp Driver, to 60 V, 0.5 A	TDE1767A	Thomson-CSF		
Driver, 8-Channel Current-Sink Driver	UDN-2595A	Sprague	45					Relay and Lamp Driver, to 65 V, Sinks 300 mA, OR Input for 48 V Telephone Relays	DS1686	† National	145		
Driver, 10-Bit, High-Voltage, High-Current	S4534	AMI			MOS Clock Driver	MH0007	† National			DS1687		† National	
Driver, 10-Bit Serial-In Latched Source	UCS-4810H	Sprague				MH0007C	National			DS3686		National	
Driver, 32-Bit for Displays, Relays, Solenoids, Print Heads and Motors	S4521	AMI	50		MH0012	† National							
	S4535	AMI			MOS Dynamic Memory, Address Refresh Logic Circuitry	MC8505	Motorola						
Dynamic RAM Controller Interface Circuit	DP84312	National											
Dynamic RAM Programmable Refresh Timer	DP84300	National	55	MOS Dynamic Memory Controller	AM2968	AMD							
					<b>MC74F2968</b>	<b>Motorola (822)</b>							
				MOS Dynamic Memory Interface, Microprocessor to 16 K RAM	MC3480	Motorola							
				MOS Dynamic Memory, 4 K Address Multiplexer and Refresh Counter	96LS32	† Fairchild							
					96LS32C	Fairchild							
					MC3232A	Motorola							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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INTERFACE—Memory and Peripheral Drivers (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Memory &amp; Peripheral Drivers (Cont'd)</b>				Dual Inverter, to 40 V, 2 A LPD4105 Lambda				Dual Peripheral AND Driver (Cont'd)			
Relay Driver, to 65 V, Sinks 300 mA, OR Input for 48 V Telephone Relays (Cont'd)				Dual Memory Driver, 400 mA Sink (for magnetic memories) MC4043 Motorola				SN75471 TI SN75476 TI			
DS3687 National				Dual MOS Clock Driver				Dual Peripheral AND Driver, for CMOS DS1631 † National DS3631 National			
Relay Driver, to 70 V, 500 mA CSR301 TeledyneC				0026 Fairchild				Dual Peripheral AND Driver, to 70 V 300 mA SN75446 TI			
Relay Driver, 5 Channel, to 65 V, 70 mA ITT7163 ITT ITT7164 ITT				MMH0026C Motorola DS0025 † National DS0026 † National DS0026C National DS0056 † National DS0056C National MH0009 † National MH0009C National MH0013 † National MH0013C National SN55369 † TI SN75369 TI				Dual Peripheral AND Driver, to 80 V 300 mA DS1611 † National DS3611 National UDN-3611M Sprague UDN-5711M Sprague UDS-3611H † Sprague UDS-5711H † Sprague			
Solenoid Driver MC3484V2 Motorola MC3484V4 Motorola				45				50			
Thermal Print Head Driver SN75270 TI SN75490 TI SN75590 TI				5				55			
Triac Controller TEA1511 Thomson-CSF				10				Dual Peripheral and Power Drivers UDN-5732M Sprague UDN-5742M Sprague			
Winchester Disc Memory Fault Detector SSI103 SiliconSys				Dual MOS Clock Driver, Bootstrapped for Single Supply Systems DS1642 National DS1671 National DS1672 National				Dual Peripheral NAND Driver 55452B † Fairchild 75452A Fairchild 75452B Fairchild 75462 Fairchild 75472 Fairchild HD75452 Hitachi DS55452 † National DS55462 † National DS75452 National DS75462 National SG55452 † SiliconG SG55462 † SiliconG SG75452 SiliconG SG75462 SiliconG SN55452B † TI SN55462 † TI SN55472 † TI SN75402 TI SN75407 TI SN75412 TI SN75432 TI SN75447 TI SN75452B TI SN75462 TI SN75472 TI			
Winchester Disc Memory Head Read/Write Circuit, for thin film heads SSI114 SiliconSys (4208)				Dual NAND Driver, HNIL, 250 mA, Open Collector 392A/C TeledyneS				60			
Winchester Disc Memory Head Selector SSI102 SiliconSys				Dual NOR Driver, HNIL, 250 mA, Open Collector 394A/C TeledyneS				65			
Winchester Disc Memory Read/Write Circuit SSI104 SiliconSys (4208)				Dual OR Driver, HNIL, 250 mA, Open Collector 393A/C TeledyneS				70			
Winchester Disc Memory Video Amplifier, for Magnetic Servo Head SSI101A SiliconSys (4208)				Dual Peripheral AND Driver 55450B † Fairchild 75450B Fairchild 75451A Fairchild 75451B Fairchild 75461 Fairchild 75471 Fairchild HD75450A Hitachi HD75451A Hitachi DS55450 † National DS55451 † National DS55461 † National DS75450 National DS75451 National DS75461 National PBD3513 RIFA SG55450 † SiliconG SG55450B † SiliconG SG55451 † SiliconG SG55460 † SiliconG SG55460B † SiliconG SG55461 † SiliconG SG75450 SiliconG SG75450B SiliconG SG75451 SiliconG SG75460 SiliconG SG75461 SiliconG				75			
Winchester Disc Memory Video Amplifier, for Thin Film Magnetic Heads SSI116 SiliconSys (4208)				Dual Peripheral AND Driver 55450B † Fairchild 75450B Fairchild 75451A Fairchild 75451B Fairchild 75461 Fairchild 75471 Fairchild HD75450A Hitachi HD75451A Hitachi DS55450 † National DS55451 † National DS55461 † National DS75450 National DS75451 National DS75461 National PBD3513 RIFA SG55450 † SiliconG SG55450B † SiliconG SG55451 † SiliconG SG55460 † SiliconG SG55460B † SiliconG SG55461 † SiliconG SG75450 SiliconG SG75450B SiliconG SG75451 SiliconG SG75460 SiliconG SG75461 SiliconG				80			
Winchester Disk Memory Read/Write Circuit MB4111 Fujitsu MB4112 Fujitsu SSI105 SiliconSys SSI108 SiliconSys (4208) SSI114 SiliconSys SSI115 SiliconSys (4208)				20				85			
Winchester, Read/Write Circuit for thin-film heads MC4290 NEC (3549)				25				90			
Winchester Read/Write Circuit, 6-Channel SSI117 SiliconSys (4208)				30				95			
Dual AND Driver, HNIL, 250 mA, Open Collector 391A/C TeledyneS				35				100			
Dual AND TTL to MOS Driver (NMOS memory interface) 9643 Fairchild SN55363 † TI SN75322 TI SN75363 TI				30				Dual Peripheral NAND Driver, for CMOS DS1632 † National DS3632 National			
Dual Buffer, to 40 V, 2 A LPD4106 Lambda				Dual CCD Memory Driver, with Enable Inputs SN75430 TI				Dual Peripheral NAND Driver, to 70 V, 300 mA MC1472 Motorola (3064) UDN-5722M Sprague SN75477 TI			
Dual CCD Memory Driver, with Enable Inputs and Protect SN75431 TI				Dual Channel ECL to MOS Driver (MOS memory interface) MC75358 Motorola				Dual Peripheral NAND Driver, to 70 V, 500 mA SN75417 TI			
Dual CMOS or TTL Driver/Translator, up to 30 V IH6201C Intersil IH6201M † Intersil				Dual Darlingington Switch, to 80 V, 1.5 A ULN-2061M Sprague ULN-2062M Sprague				Dual Peripheral NAND Driver, to 80 V, 300 mA DS1612 † National DS3612 National UDN-3612M Sprague UDN-5712M Sprague UDS-3612H † Sprague UDS-5712H † Sprague			
Dual ECL to TTL/MOS Driver SN75441 TI				Dual Half Bridge Driver, 1.5 A SG3636 SiliconG				Dual Peripheral NOR Driver 75474B Fairchild HD75454 Hitachi DS55454 † National DS55464 † National DS75454 National DS75464 National SG55454 † SiliconG SG55464 † SiliconG			
								(Continued)			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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**INTERFACE-Memory and Peripheral Drivers (Cont'd)**

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line					
<b>Memory &amp; Peripheral Drivers (Cont'd)</b>																
Dual Peripheral NOR Driver (Cont'd)	SG75454	SiliconG	5	Dual Power MOS Driver	<b>ICL7667</b>	<b>Intersil (3013)</b>	60	Quad Darlington Switch, to 80 V, 1.5 A (Cont'd)	ULS-2068H	† Sprague	115					
	SG75464	SiliconG			Dual Power MOSFET Driver	TSC450C		TeledyneS				ULS-2069H	† Sprague			
	SN55454B	† TI				TSC450M		† TeledyneS				ULS-2070H	† Sprague			
	SN55464	† TI			Dual TTL to MOS Driver (MOS memory interface)	DS75361		National	65			ULS-2071H	† Sprague			
	SN55474	† TI				DS75362		National					ULS-2074H	† Sprague		
	SN75404	TI				MH8805		National					ULS-2075H	† Sprague		
	SN75407	TI				SN75350		TI					ULS-2076H	† Sprague		
	SN75414	TI				SN75361A		TI					ULS-2077H	Sprague		
	SN75434	TI		10	Dual 2-Input AND Power Driver (40 V, 2 A)	LPD4101		Lambda	70			SN75065	TI	120		
	SN75449	TI				Dual 2-Input NAND Power Driver (40 V, 2 A)		LPD4102		Lambda					SN75067	TI
	SN75454B	TI				Dual 2-Input NOR Power Driver (40 V, 2 A)		LPD4103		Lambda					SN75069	TI
	SN75464	TI				Dual 2-Input OR Power Driver (40 V, 2 A)		LPD4104		Lambda					ULN2064	TI
	SN75474	TI				Dual 4 A Solenoid Driver		UDN-2975W		Sprague					ULN2065	TI
	SN75479	TI						UDN-2976W		Sprague					ULN2066	TI
Dual Peripheral NOR Driver, for CMOS	DS1634	† National	15	Dual 4-Input AND Driver, HNIL 250 mA Open Collector	390A/C	TeledyneS	75	Quad Darlington Switch, to 80 V, 4 A	<b>UDM-2879</b>	<b>Sprague (3088)</b>	125					
	DS3634	National			Dual 4-Input NAND Driver, HNIL, 250 mA Open Collector	395A/C		TeledyneS		Quad Darlington Switch, to -50 V, 1.75 A		SG2841	SiliconG			
Dual Peripheral NOR Driver, to 30 V, 300 mA	SN75479	TI	20	Quad Bipolar Driver	SG293A	SiliconG	75	Quad Darlington Switch	TEB1013	Thomson-CSF	130					
Dual Peripheral NOR Driver, to 30 V, 500 mA	SN75419	TI			Quad Darlington Driver, to 100 V, 1.5 A	FT5753		Fujitsu		Quad Driver for Automotive Applications		DS3656	National			
Dual Peripheral NOR Driver, to 80 V, 300 mA	DS1614	† National			Quad Darlington Driver, to 100 V, 3 A	FT5754		Fujitsu		Quad ECL to MOS Clock Driver		HD2923	Hitachi			
	DS3614	National			Quad Darlington Driver, to 100 V, 5 A	FT5755		Fujitsu		Quad Fault Protected Peripheral Driver		DS3668	National			
	UDN-3614M	Sprague			Quad Darlington Switch, to 50 V, 1.5 A	SN75064		TI	80	Quad High Current Peripheral Driver		DS3658	National	135		
	UDN-5714M	Sprague			Quad Darlington Switch, to 50 V, 4 A	SN75066		TI					DS3669		National	
	UDS-3614H	† Sprague		Quad Darlington Switch, to 80 V, 1.5 A	SN75068	TI				SN75436	TI					
Dual Peripheral OR Driver	75453A/B	Fairchild	25	Quad Darlington Switch, to 50 V, 1.5 A	SN75064	TI	85	Quad Latch/Driver	MD121	AnalogSys	140					
	HD75453	Hitachi			Quad Darlington Switch, to 50 V, 4 A	UDN-2878		<b>Sprague (3088)</b>				UCN-4401A	Sprague			
	MC75453	Motorola			Quad Darlington Switch, to 80 V, 1.5 A	SG2064		SiliconG		Quad MOS Clock Driver		DS3245	National			
	DS55453	† National				SG2065		SiliconG		Quad, MOS Memory Decoder/Clock Driver		DS36143	National			
	DS55463	† National				SG2066		SiliconG		Quad MOS Memory I/O Register		DS16147	† National			
	DS75453	National		30		SG2067		SiliconG				DS16177	† National			
	DS75463	National						SG2068	SiliconG				DS1647	† National		
	SG55453	SiliconG						SG2069	SiliconG				DS1677	† National		
	SG55463	SiliconG		35		SG2070		SiliconG				DS36147	National	150		
	SG75453	SiliconG						SG2071	SiliconG				DS36177		National	
	SG75463	SiliconG						SG2072	SiliconG				DS36177		National	
	SN55453B	† TI						SG2073	SiliconG				DS3647		National	
	SN55463	† TI						SG2074	SiliconG				DS3677		National	
	SN55473	† TI						SG2075	SiliconG							
	SN75403	TI	40		SG2076	SiliconG					155					
	SN75408	TI				SG2077	SiliconG									
	SN75413	TI				ULN-2064B	Sprague	100	Quad Multiplexer/Driver, for MOS Systems	DS1648		† National				
	SN75433	TI				ULN-2065B	Sprague					DS1678	† National			
	SN75448	TI				ULN-2066B	Sprague					DS3648	National			
	SN75453B	TI				ULN-2068B	Sprague					DS3678	National			
	SN75463	TI	45		ULN-2069B	Sprague			Quad NAND Gate 60 V Power Driver	UDN-2541	Sprague					
	SN75473	TI				ULN-2070B	Sprague			Quad NAND Gate 80 V Power Driver	UDN-2542	Sprague				
	SN75478	TI			ULN-2071B	Sprague			Quad NAND TTL to MOS Driver (MOS memory interface-clock driver)	HD2912	Hitachi	160				
Dual Peripheral OR Driver, ECL Input	SN75441	TI	50		ULN-2074B	Sprague				HD2916	Hitachi					
Dual Peripheral OR Driver, for CMOS	DS1633	† National				ULN-2075B	Sprague				DS75365		National			
	DS3633	National			ULN-2076B	Sprague				3207A	Signetics					
Dual Peripheral OR Driver, to 30 V, 300 mA	SN75478	TI			ULN-2077B	Sprague				3207A-1	Signetics					
Dual Peripheral OR Driver, to 30 V, 500 mA	SN75418	TI	55		ULS-2064H	† Sprague	110			SN55355	† TI					
Dual Peripheral OR Driver, to 80 V, 300 mA	DS1613	† National				ULS-2065H			† Sprague				(Continued)			
	DS3613	National				ULS-2066H			† Sprague							
	UDN-3613M	Sprague				ULS-2067H		† Sprague								
	UDN-5713M	Sprague														
	UDS-3613H	† Sprague														
	UDS-5713H	† Sprague														

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

INTERFACE—Memory and Peripheral Drivers (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Memory &amp; Peripheral Drivers (Cont'd)</b>				Quad 2-Input NAND Driver, to 30 V, sinks 250 mA				Octal High-Voltage Driver for Electrostatic Printers			
				MC693 Motorola				DH0069 National 110			
Quad NAND TTL to MOS Driver (MOS memory interface-clock driver) (Cont'd)				Quad 2-Input NAND Power Driver, Open Collector, to 100 V, Sinks 500 mA				Octal Latched Peripheral Driver			
SN55365 † TI				UHC/D-407 † Sprague				DP7310 National			
SN75365 TI				UHC/D-408 † Sprague				DP7311 National			
Quad Negative Voltage Relay Driver				UHC/D-507 † Sprague				DP8310 National			
DS3680 National				UHC/D-508 † Sprague				DP8311 National			
DS3680 TI				UHP-407 Sprague				Octal MOS Driver, Three-State			
Quad NMOS Memory Driver				UHP-408 Sprague				DP84240 National 115			
MC3459 Motorola				UHP-507 Sprague				DP84244 National			
MC3460 Motorola				UHP-508 Sprague				DS1628 † National			
DS1644 † National				Quad 2-Input NOR Driver, to 70 V, Sinks 300 mA				DS3628 National			
DS1674 † National				UDN-5733M Sprague				Two Phase Oscillator/Clock Driver (for MOS systems)			
DS36144 National				UDS-5733H † Sprague				DS7803 † National			
DS3644 National				Quad 2-Input NOR Power Driver, Open Collector, to 100 V, Sinks 500 mA				DS7807 † National			
DS3674 National				UHC/D-432 † Sprague				DS8803 National			
Quad PIN Diode Driver (also see: linear-other devices)				UHC/D-433 † Sprague				DS8807 National			
UDS-5790H † Sprague				UHC/D-532 † Sprague				Eight Latch/Drivers			
UDS-5791H † Sprague				UHC/D-533 † Sprague				UCN-4801A Sprague			
Quad Port Driver, for 5270 RAM				UHP-432 Sprague				4-Input AND Current Driver (45 V, to 1.5 A peak)			
DS1640 † National				UHP-433 Sprague				DH0006 † National			
DS1670 † National				UHP-532 Sprague				DH0006C National			
Quad Power NAND Driver				UHP-533 Sprague				4-Input AND High Voltage-High Current Driver (4.5 V, to 3 A peak)			
CA3219 RCA (3662)				Quad 2-Input OR Driver, to 70 V, Sinks 300 mA				DH0008 † National			
Quad Power Peripheral Driver				UDN-5703A Sprague				4-Input NAND High Voltage, High Current Drivers (40 V, 0.15 to 0.25 A)			
7254 Intel				UDS-5703H † Sprague				DH0011 † National			
VQ1000 Siliconix				Quad 2-Input OR Power Driver, Open Collector, to 100 V, Sinks 500 mA				4-Input NAND High Voltage, High Current Drivers (50, 70 or 100 V, 0.25 to 0.5 A)			
VQ1001 Siliconix (3085)				UHC/D-402 † Sprague				DH0017C National			
VQ1004 Siliconix (3085)				UHC/D-403 † Sprague				DH0018C National			
VQ1006 Siliconix (3085)				UHC/D-502 Sprague				6-Bit MOS Refresh Counter/Driver			
VQ2001 Siliconix (3085)				UHC/D-503 Sprague				DS1646 † National			
VQ2004 Siliconix (3085)				UHP-402 Sprague				DS1676 † National			
VQ2006 Siliconix (3085)				UHP-403 Sprague				8-Bit Serial Input, Latched Sink Driver			
VQ3001 Siliconix (3085)				UHP-502 Sprague				UCN-4821A Sprague			
VQ7254 Siliconix (3085)				UHP-503 Sprague				UCN-4822A Sprague			
VC01 Supertex				Hex Driver, CMOS/TTL (line, LED, Relay Driver)				UCN-4823A Sprague			
VC02 Supertex				MD-210 AnalogSys				UCS-4821H † Sprague			
VC13 Supertex				Hex Inverter/MOS Driver, Disable Causes Logic 1 State				UCS-4822H † Sprague			
VQ1000 Supertex				DS16149 † National				UCS-4823H † Sprague			
VQ7254 Supertex				DS16179 † National							
Quad Power Strobe				DS36149 National							
HD6600-2 † Harris				DS36179 National							
HD6600-5 Harris				<b>SN548436 † TI (1035,1036)</b>							
HD6600 National				<b>SN548437 † TI (1035,1036)</b>							
Quad Predriver, Open Collector, 50 mA Sink (for magnetic memories)				<b>SN748436 TI (1035,1036)</b>							
MC4042 Motorola				<b>SN748437 TI (1035,1036)</b>							
Quad TTL to MOS Driver, Three-State				Hex Inverter/MOS Driver, Three-State							
SN75367 TI				DS1649 † National							
Quad TTL to NMOS Memory Driver (for 2105, 2107, etc.)				DS1679 † National							
3245 Fairchild				DS3649 National							
9645 Fairchild				DS3679 National							
Quad TTL to NMOS Memory Driver (for 7001, etc.)				Hex Latch/Driver, for MOS Memories							
MC3466 Motorola				DS1645 † National							
Quad 2-Input AND Driver (to 70 V, sinks 300 mA)				DS1675 † National							
UDN-5706A Sprague				DS3645 National							
UDS-5706H † Sprague				DS3675 National							
Quad 2-Input AND Power Driver, Open Collector (to 100 V, sinks 500 mA)				Octal Dynamic Memory Driver, Three-State							
UHC/D-400 † Sprague				AM2965C AMD							
UHC/D-406 † Sprague				AM2965M † AMD							
UHC/D-500 Sprague				AM2966C AMD							
UHC/D-506 Sprague				AM2966M † AMD							
UHP-400 Sprague				Octal Dynamic RAM Driver, Three-State							
UHP-406 Sprague				SN548700 † MNM (804,819)							
UHP-500 Sprague				SN548730 † MNM (804,819)							
UHP-506 Sprague				SN548731 † MNM (804,819)							
Quad 2-Input NAND Driver (for 70 V, sinks 300 mA)				SN548734 † MNM (804,819)							
UDN-5707A Sprague				SN748700 MNM (804,819)							
UDS-5707H † Sprague				SN748730 MNM (804,819)							
				SN748731 MNM (804,819)							
				SN748734 MNM (804,819)							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



# IC MASTER

## INTERFACE—Sense Amplifiers

Function	Device	Source	Line	Function	Device	Source	Line	
<b>Sense Amplifiers</b>				<b>Dual Sense Amplifier (for MOS memory or line receiver)</b>				
MOS to TTL Level Converter, High Speed, Three-State				(Cont'd)				
	MC4000	Motorola			SN75208	TI		
	MC4300	† Motorola			DS3604	National	55	
Translators: See also Digital-TTL (Translators)					DS1603	† National		
Dual Core Memory Sense Amplifier, Complementary Output, Latch Capability					DS3603	National		
	55S20	Fairchild		<b>Dual Sense Amplifier (NMOS memories to ECL 10K)</b>				
	75S20	† Fairchild	5		HD103461	Hitachi		
	SG5520	† SiliconG		<b>Triple Current Sense Latch</b>				
	SG5521	† SiliconG			SG3551	SiliconG		
	SG7520	SiliconG		<b>Quad Sense Amplifier, Three-State</b>				
	SG7521	SiliconG			MC3430	Motorola	60	
	SN5520	† TI	10		MC3431	Motorola		
Dual Core Memory Sense Amplifier/Data Register					MC3432	Motorola		
	SG55236	† SiliconG			MC3433	Motorola		
	SG75236	SiliconG			DS1651	† National	65	
	SN55236	† TI			DS1653	† National		
	SN55237	† TI			DS3651	National		
Dual Core Memory Sense Amplifier, Separate Inverted Outputs					DS3653	National		
	55S234	Fairchild	15	<b>Hex MOS Sense Amplifier (MOS to TTL Converter)</b>				
	75S234	† Fairchild		Three-State				
	SN55234	† TI			DS3605	National	70	
	SN75234	TI			DS3606	National		
Dual Core Memory Sense Amplifier, Separate Inverted Outputs, Test Points					DS3607	National		
	75239	Fairchild			DS3608	National		
Dual Core Memory Sense Amplifier, Separate Open Collector Outputs					<b>Octal Core Memory Driver</b>			
	SG5534	† SiliconG	20		SN55329	† TI		
	SG5535	† SiliconG		4-Input Sense Amplifier (for plated wire or thick/thin film memories)				
	SG7534	SiliconG			MC1444	Motorola		
	SG7535	SiliconG			MC1544	† Motorola		
	SN55232	† TI						
Dual Core Memory Sense Amplifier, Separate Open Collector Outputs, Test Points								
	SG5538	† SiliconG	25					
	SG5539	† SiliconG						
	SG7538	SiliconG						
	SG7539	SiliconG						
Dual Core Memory Sense Amplifier, Separate Outputs								
	55S24	Fairchild	30					
	75S24	† Fairchild						
	HA1902	Hitachi						
	SG5524	† SiliconG						
	SG5525	† SiliconG						
	SG7524	SiliconG						
	SG7525	SiliconG	35					
	SN5524	† TI						
Dual Core Memory Sense Amplifier, Separate Outputs, Test Points								
	SG5528	† SiliconG	40					
	SG5529	† SiliconG						
	SG7528	SiliconG						
	SG7529	SiliconG						
	SN5528	† TI						
Dual Core Memory Sense Amplifier, Single Open Collector Output								
	SG5522	† SiliconG	45					
	SG5523	SiliconG						
	SG7522	SiliconG						
	SG7523	SiliconG						
	SN5522	† TI						
Dual Formatter/Sense Amplifier for Bubble Memories								
	7242	Intel						
Dual MOS to TTL Level Converter, Latch, Three-State (Sense Amp)								
	MC74468	Motorola						
	N8T25	SiliconG						
Dual Sense Amplifier (for MOS memory or line receiver)								
	DS75207	National	50					
	DS75208	National						
	SN75207	TI						

(Continued)

† Military Temperature Range (−55° to 125°C)

\* Typical Value

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INTERFACE-Transmitters-Receivers

Function	Max Serial Data Rate in KHZ	Supply Voltage, V	Device	Source	Line	Function	Max Serial Data Rate in KHZ	Supply Voltage, V	Device	Source	Line
<b>Serial Transmitters- Receivers</b>						Baud Rate Generator (programmable divider), Dual 1000 5 (Cont'd)					
(ACIA) Asynchronous Communications Interface Adapter (Links 8-Bit bidirectional data bus to serial asynchronous data communications, including to 6860)	50	5	S6551 S6551A S6850 F6850 MC6850 EF6850	AMI (1449) AMI (1449) AMI Fairchild (1463) Motorola (1505) Thomson-CSF	5				AY5-8116T AY5-8136 AY5-8136T COM8116 COM8116T COM8136 COM8136T WD1945	GI GI GI SMC SMC SMC SMC Western	55
Advanced Data Link Controller	2 Mb/s	5	MC854	Motorola				12.5	COM5016 COM5016T COM5036 COM5036T	SMC SMC SMC SMC	60
ARINC-429 Buffer Receiver 1-15			EF4440	Thomson-CSF							
ARINC-429 Receiver/Transmitter	100	5	HS3182 HS3282 EF4442 WD1993-01 WD1993-02 WD1993-03	† Harris † Harris (4648) Thomson-CSF Western Western Western	10	BOART (Bus Orientated Programmable Asynchronous Receiver/Transmitter)	9.6	5	TR1983	Western	
ASTRO (asynchronous/synchronous receiver/transmitter) to Interface Serial Communications Channel with a Parallel Digital System (i.e. microprocessors)	1000	12.5 12, ±5	WD1931 COM1671 UC1671	Western SMC Western	15	Bus Interface Circuit (MIL-STD-1553B)	5000	5	HS3273	† Harris	65
Asynchronous Addressable Receiver/Transmitter	—	4.75-11.5	MMS4240 MC14469	National Motorola		Command/Response Manchester II Converter (MIL-STD-1553B)	1000	5	BUS-8937 SSM-2012	† DDC SSM	
Asynchronous Communications Element	56	5	WD8250	Western	20	Digital Modem (Modulation, demodulation and supervisory control, up to 600 BPS.)	0.6	5	MC6860	Motorola (1505,3536)	
Asynchronous Serial Interface Component			MA2232	National (2049)		DUART (Dual Universal Asynchronous Receiver and Transmitter). Two-channel UART, baud-rate generator, 16-Bit counter/timer, I/O ports.	1000	5	26810 26814 26818 SCN2681 SCN68681	Signetics Signetics Signetics Signetics (1675) Signetics (1694)	70
Asynchronous Serial Manchester Adapter	1000	5	HD6408	Harris (1492,1496)		EPCI (Enhanced Programmable Communications Interface) Serial/Parallel receiver/transmitter-Synchronous and asynchronous with baud rate generator.			MC2261A MC2261B MC2261C	Motorola Motorola Motorola	75
Baud Rate Generator (programmable divider)	1000	5	COM8046 COM8046T COM8126 COM8126T COM8146 COM8146T	SMC SMC SMC SMC SMC SMC	25		1 Kb/s	5	MC68661A MC68661C	Motorola Motorola (1506,1512)	
		12.5	COM5026 COM5026T COM5046 COM5046T	SMC SMC SMC SMC	30		15.6 Kb/s	5	MC68661B SCN68661	Motorola (1506,1512) Signetics (1673)	80
		12, ±5	F4702BC F4702BM HD4702-2 HD4702-9 IM4702 IM4703 MC14411 MM5307	Fairchild † Fairchild † Harris (1492) † Harris (1492) Intersil Intersil Motorola (3536) National	35	IBM 3274/3276 Compatible COAX Receiver/Transmitter	2358	5	COM9004	SMC	
Baud Rate Generator (programmable divider), Dual 3077/19.2		5	WD1943-00 WD1943-02 WD1943-05	Western Western Western	40	Link Controller, X.25 level 2	100	5,12	WD2501-01 WD2511-01	Western Western (1174)	85
		12.5	BR1941-00 BR1941-02 BR1941-05	Western Western Western	45		500	5,12	WD2501-03 WD2511-05	Western Western (1174)	
	614/19.2	5	WD1943-04 WD1943-06	Western Western			1000	5,12	WD2501-11 WD2511-11	Western Western (1174)	90
		12.5	BR1941-04 BR1941-06	Western Western	50	Manchester Encoder-Decoder	1000	5	AM7991 MB501 HD6409-2 HD6409-9 HS15530RH HD15530-2	AMD (1433,3327) Fujitsu † Harris (1493,1497) Harris (1493,1497) † Harris (4680) † Harris (1493,1498)	95
	1000	5	AY5-8116	GI			1250	5			(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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**IC MASTER**

**INTERFACE-Transmitters-Receivers (Cont'd)**

Function	Max Serial Data Rate in KHZ	Supply Voltage, V	Device	Source	Line	Function	Max Serial Data Rate in KHZ	Supply Voltage, V	Device	Source	Line
<b>Serial Transmitters- Receivers (Cont'd)</b>						<b>Transceiver, MIL-STD-1553A/B</b>					
Manchester Encoder-Decoder	1250	5	HD15530-9	Harris (1493,1498)	40	1000	± 12	BUS-8557	† DDC	40	
			HD15531-2	† Harris		± 15/± 12	BUS-63105	† DDC			
	2500	5	HD15531B	Harris (1493,1499)			± 15,5	BUS-8553	† DDC		
MIL-STD-1553 Dumb Remote Terminal Unit (hybrid)			BUS65101	† DDC	5	Transmitter/Decoder (Bi-Phase, IBM 3270)		5	DP8340	National	
MIL-STD-1553 Encoder/Decoder			MT32008	DDC		Transmitter/Encoder (Bi-Phase)	3500	5	DP8342	National	45
MIL-STD-1553 Terminal Bit Processor			BUS-64100	DDC		UART, MIL-STD-1553A	1000	5	COM1553A	† SMC	50
MPCC (multi-protocol communications controller) Bit and Byte Oriented	1 Mb/s	5	SCN2652A	Signetics	10	UART (Universal Asynchronous Receiver-Transmitter) (complete serial to parallel and parallel to serial interface)	1.2	- 12,5	10371	Rockwell	
	2 Mb/s	5	MC2652	Motorola (1511)		20			- 12,5	TR1402	Western
			MC2652-2	Motorola (1511)	15		30	5	AY3-1015D	GI	
			MC68852	Motorola (1506,1511)					- 12,5	MM5303	National
			MC68852-2	Motorola (1506,1511)			40	5	COM1863	SMC	55
			µPD7201	NEC (1531)					COM8017	SMC	
Parallel to Serial Interface		5 V	CY232	Cybernetic					COM8018	SMC	60
PCI (Programmable Communications Interface) Serial/parallel receiver/transmitter—synchronous and asynchronous with baud rate generator.	1000 b/s	5	SCN2651C	Signetics (1669)				- 12,5	COM8502	SMC	
PKCC (Programmable Keyboard and Communications Controller) UART, baud rate generator and keyboard encoder	1000 b/s	5	SC2671ACS	SiliconG			56	5	MS8250	Hitachi (1514)	65
									WD2123	Western	
PSAR (Programmable Synchronous-Asynchronous Receiver) Synchronous/Asynchronous serial to parallel converter with programmable character length and programmable serial data rate.	100	- 12,5	PT1472B	Western	20		60	5	TR1863-00	Western	
	640	- 12,5	PT1472B-01	Western						TR1865-00	Western
		- 12, ± 5	NC2259	Nitron			100	5	82C52	† Harris (1482,1494)	70
PSART (Programmable Synchronous-Asynchronous Receiver-Transmitter) Serial to parallel and parallel to serial converter that can operate in Full Duplex Mode.	50	5	8251	Intel					HD6406	† Harris (1492,1495)	
			µPD8251	NEC					CDP1854AC	† RCA (1595)	75
PSAT (Programmable Synchronous-Asynchronous Transmitter) Synchronous/Asynchronous parallel to serial converter that has programmable character length and programmable serial data rate.	100	- 12,5	PT1482B	Western					HD6402C-9	Harris (1492)	
	200	- 12, ± 5	NC2257	Nitron	25				TR1863-02	Western	80
	640	- 12,5	NC2260	Nitron						TR1865-02	
Receiver/Decoder (Bi-Phase)	3500	5	DP8343	National					HD6402-2	† Harris (1492)	85
Receiver/Decoder (Bi-Phase, IBM 3270)			DP8341	National					HD6402-9	Harris (1492)	
SDLG/HDLC/ADCCP Controller	500	5	WD193X-00/10	Western	30	Universal Communications Interface (Receives or transmits data to serial data bus when addressed and commanded by bus. Links the bus to serial or parallel I/O devices.)	500	- 10,5	UMC-16	Trans-Data	85
	1000	5	WD193X-01/11	Western						BUS65201	
	1500	5	WD193X-02/12	Western		Universal Data Bus Transceiver (hybrid)			BUS63102	† DDC	
	2000	5	WD193X-03/13	Western		Universal Digital-Loop Transceiver	80	5	MC145422	Motorola (3538)	90
SPCC (Sync-Protocol Communications Controller) Bit and Byte Oriented	1000	5	F3846	Fairchild					MC145423	Motorola	
			F6856	Fairchild (1460)	35				MC145426	Motorola (3538)	
Synchronous Receiver/Transmitter (Bi-Sync/SDLG)	800	± 5, 12	µPD379	NEC		USRT (universal synchronous receiver-transmitter) Complete serial to parallel and parallel to serial interface.	250	- 12,5	COM2601	SMC	
Synchronous Serial Data Adaptor	2 Mb/s	5	MC6852	Motorola (1505)			500	5	S2350	AMI	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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INTERFACE-Transmitters-Receivers (Cont'd)

Function	Max Serial Data Rate in kHz	Supply Voltage, V	Device	Source	Line
<b>Serial Transmitters- Receivers</b>			<b>(Cont'd)</b>		
USYNRT (universal synchronous receiver/transmitter) Multi-Protocol, Bit and Byte Oriented	1500	5, 12	COM5025	SMC	
Dual Channel Asynchronous Serial Interface Circuit			MC2681	Motorola	
			<b>MC68681</b>	<b>Motorola</b> (1506.1512)	
			SCN68681	Signetics	
Dual Redundant MIL-STD-1553 Remote Terminal Unit (hybrid)			BUS65122	† DDC	5

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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# ABBREVIATIONS OF COMPANY NAMES

**Action Ins** Action Instruments  
**AD** Analog Devices  
**ADT** Advanced Digital Technology  
**Advent** Advent Products, Inc.  
**Alphatron** Alphatron  
**AMA** American Automation  
**AMD** Advanced Micro Devices  
**AMI** American Microsystems, Inc.  
**Amperex** Amperex Electronic Corp.  
**Analogic** Analogic  
**Analog Sys** Analog Systems  
**APC** Applied Micro Circuits  
**Apex** Apex Microtechnology  
**APM** Applied Microsystems Corp.  
**Appl Sys** Applied Systems Corp.  
**APT** Applied Microtechnology  
**Aptek** Aptek Microsystems  
**Array Tech** Array Technology  
**AWI** AWI Electronics

**Barvon** Barvon Research  
**Bedford** Bedford Computer Systems Inc.  
**Burr-Brown** Burr-Brown

**CAE** Computer Aided Engineering  
**Cal Devices** California Devices  
**Cermetek** Cermetek  
**CGRS** CGRS Microtech Inc.  
**Cherry** Cherry Semiconductor  
**CIC** Custom Integrated Circuits  
**CirTech** Circuit Technology  
**Citel** Citel  
**Comlinear** Comlinear Corporation  
**CMA** Custom MOS Arrays  
**Comark** Comark Corp.  
**Comdial** Comdial Semiconductor  
**Comp Auto** Computer Automation  
**Compas** Compas Microsystems  
**Cont Logic** Control Logic Inc.  
**Control Sys** Control Systems Microsystems Div.  
**CreMicro** Creative Micro Systems  
**Cromemco** Cromemco, Inc.  
**Cubit** Cubit Inc.  
**Curtis** Curtis Electro Devices, Inc.  
**Cybernetic** Cybernetic Micro Systems  
**Cybersys** Cybersystems  
**Cybertek** Cybertek Inc.

**Data General** Data General  
**Data I/O** Data I/O  
**Data Trans** Data Translation  
**Datel** Datel  
**Datricon** Datricon Corporation  
**DDC** Data Devices Corporation  
**DEC** Digital Equipment Corporation  
**Die-Tech** Die-Tech  
**Digelec** Digelec Corp.  
**Digitel** Digitel, Inc.  
**Dionics** Dionics Inc.  
**Dist Comp** Distributed Computer Systems  
**Divers Tech** Diversified Technology

**E-HI** E-H International, Inc.  
**EDI** Electronic Designs Inc.  
**Elind** Elind Elettronica Industriale  
**EMM-SESCO** EEM-SESCO  
**Emulogic** Emulogic Inc.  
**ETI Micro** ETI Micro  
**Exar** Exar Integrated Systems  
**Exel** Exel Microelectronics

**Fairchild** Fairchild  
**Ferranti** Ferranti Electric  
**Force** Force Computers  
**Fujitsu A** Fujitsu America  
**Fujitsu** Fujitsu Microelectronics, Inc.

**GI** General Instrument  
**GTE Micro** GTE Microcircuits

**Harris** Harris Semiconductor  
**Heurikon** Heurikon Corp.  
**Hilevel** Hilevel Technology, Inc.  
**Hitachi** Hitachi America, Ltd.  
**Holt** Holt Inc.  
**HP** Hewlett-Packard  
**Hughes** Hughes Aircraft, Solid State Products  
**Hybrid Sys** Hybrid Systems  
**HyComp** HyComp

**ICC** International Cybernetics  
**IDT** Integrated Device Technology  
**IMI** International Microcircuits, Inc.  
**IMP** International Microelectronic Products  
**IMS** Industrial MicroSystems Inc.  
**Infosphere** Infosphere  
**Inmos** Inmos  
**IntCirEng** Integrated Circuit Engineering  
**IntCirSys** Integrated Circuit Systems  
**IntCompSys** Integrated Computer Systems  
**int Tech** Integrated Technology Corp.  
**Intech** Intech Microcircuits Div.  
**Intel** Intel  
**Interdesign** Interdesign  
**Intersil** Intersil  
**Intronics** Intronics  
**ITT** ITT Semiconductors

**Kinetic Sys** Kinetic Systems  
**Kontron** Kontron Electronics

**Lambda** Lambda Semiconductor  
**Linear Tech** Linear Technology  
**LSI Comp** LSI Computer Systems  
**LSI Logic** LSI Logic Corporation

**Master Logic** Master Logic Corporation  
**Matrix** Matrix Corp.  
**Matrox** Matrox Electronic Systems  
**Maxim** Maxim Integrated Products  
**MCC** Micro-Computer Control  
**MCE** MCE Electronics  
**Micrel** Micrel  
**Micro Innov** Micro Innovators  
**Micropac** Micropac Industries  
**Micro Net** Micro Networks  
**Micro Pwr** Micro Power Systems  
**Micro Sci** Micro Sciences Corp.  
**Micro-Link** Microcircuits Technology  
**Micro-Link** Micro-Link Corporation  
**Micron** Micron Technology  
**MillerTron** MilerTronics  
**Miller** Miller Technology  
**Mitel** Mitel Semiconductor  
**Mitsubishi** Mitsubishi Electronics  
**MMI** Monolithic Memories, Inc.

**Mostek** Monolithic Systems Corp.  
**Motorola** Mostek  
**MRC** Motorola Semiconductor  
**Murray** MRC Systems  
**Monosil** Murray Consulting

**National** National Semiconductor  
**NCM** NCM Corp.  
**NCR** NCR Corp., Microelectronics Division  
**NEC** NEC Electronics  
**Nitron** Nitron

**DAE** Oliver Advanced Engineering  
**Octagon** Octagon Systems Corp.  
**OEI** Optical Electronics Inc.  
**Ohio Sci** Ohio Scientific  
**OKI** OKI Semiconductor  
**Omnibyte** Omnibyte Corp.  
**Onset** Onset Computer Corp.

**Panasonic** Panasonic  
**Pico Design** Pico Design  
**Polycore** Polycore Electronics  
**Plessey** Plessey Semiconductors  
**PMI** Precision Monolithics, Inc.  
**PragDes** Pragmatic Design Inc.  
**Pro-Log** Pro-Log Corp.

**Quay** Quay Corp.

**Raytheon** Raytheon Semiconductor  
**RCA** RCA Solid State Division  
**RCI Data** RCI Data  
**RELMS** Relational Memory Systems  
**Reticon** Reticon  
**RIFA** RIFA  
**Rockwell** Rockwell, Microelectronic Devices  
**RTC** Riehl Time Corporation

**Sanyo** Sanyo  
**SBE** SBE, Inc.  
**SEEQ** SEEQ Technology, Inc.  
**SPI** Semi Processes Inc.  
**Siemens** Siemens  
**Si-Fab** Si-Fab  
**Signetics** Signetics  
**SGS** SGS Semiconductor  
**Sharp** Sharp  
**Silicon G** Silicon General  
**Siliconix** Siliconix  
**Silicon Sys** Silicon Systems Inc.  
**Siltronics** Siltronics  
**SMC** Standard Microsystems Corp.  
**S MOS** S MOS Systems  
**Solarise** Solarise Enterprises  
**Solitron** Solitron Devices  
**Sprague** Sprague Electric Company  
**SSM** Solid State Micro Technology

for Music  
**SSS** Solid State Scientific  
**Stag** Stag Microsystems  
**STC** Storage Technology Corp.  
**STD** STD Microsystems  
**Struc Des** Structured Design Inc.  
**Stynetic** Stynetic Systems  
**Sunrise** Sunrise Electronics  
**Sunshine** Sunshine Semiconductor  
**Supertex** Supertex Inc.  
**Symtek** Symtek Corp.  
**Synertek** Synertek  
**Sys Innov** Systems Innovations

**Tau Zero** Tau Zero Inc.  
**Technitrol** Technitrol  
**Tektronix** Tektronix  
**Teledyne C** Teledyne Crystalonics  
**Teledyne P** Teledyne Philbrick  
**Teledyne S** Teledyne Semiconductor  
**Telefunken** Telefunken  
**Telmos** Telmos  
**Teltone** Teltone Corporation  
**TI** Texas Instruments  
**Third Domain** Third Domain  
**Thomson-CSF** Thomson-CSF Components Corp.  
**Toshiba** Toshiba America  
**Trans-Data** Trans-Data  
**TRW** TRW LSI Products

**Unitrode** Unitrode  
**Universal** Universal Semiconductor, Inc.

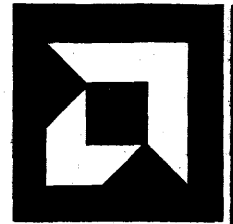
**Varix** Varix Corp.  
**VLSI Design** VLSI Design Associates  
**VTI** VLSI Technology, Inc.  
**Votrax** Votrax

**Weitek** Weitek Corporation  
**Western** Western Digital  
**Wintek** Wintek Corp.

**Xicor** Xicor, Inc.  
**Xycom** Xycom

**Zendex** Zendex Corp.  
**Zilog** Zilog  
**ZyMOS** ZyMOS Corporation  
**Zytrex** Zytrex Corp.

# Advanced Micro Devices



## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family

### DISTINCTIVE CHARACTERISTICS

- Standard I/O pin-out organization
- 48mA Commercial  $I_{OL}$
- 32mA Mil  $I_{OL}$
- Standard clock and output enable pin-outs
- 24-pin slim 0.3 inch wide DIP package
- IMOX Speed
  - 7.5ns typical CP to Y for registers
  - 4.0ns typical D to Y for buffers
- Wide data paths and flexible control
  - 10 bits for video or wide addresses
  - 9 bits for byte plus parity buses
  - 8-bit with multiple enables
- 100% Product assurance screening to MIL-STD-883 requirements

### FUNCTIONAL DESCRIPTION

The Am29800 Family provides a completely standardized functional family of registers, latches, buffers, transceivers and parity check-and-regenerate functions optimized for bus interface applications. Each is packaged in the standard 24-pin x 0.3" wide DIP package to allow LSI functionality in the minimum board area. Board layout is eased by the standardization of inputs on the left and outputs on the right, directly across from each other. Output drive levels are standardized at 48mA Commercial and 32mA Military.

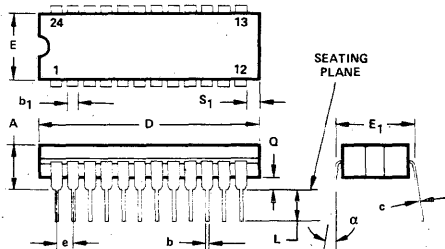
All functions are implemented in AMD's new proprietary IMOX™ (Implanted Micro OXide) process to provide the optimum in speed power product. Typical benchmark speeds are 7.5ns typical CP-to-Y for registers, and 4.0ns typical D-to-Y for buffers.

The basic 29800 Family functions are available in 10-bit, 9-bit and 8-bit configurations with broad control flexibility aimed at minimizing the SSI/MSI content of LSI systems. The 10-bit devices make it easy to interface data plus controls or for 2 parts to interface 20-bit address lines. The 9 bits function are ideal for byte plus parity bus structures. The parity check-and-regenerate functions are designed for interfacing non-parity peripherals to parity organized buses.

All of the functional types have the pin-out format shown below.

### PHYSICAL DIMENSIONS Dual-In-Line

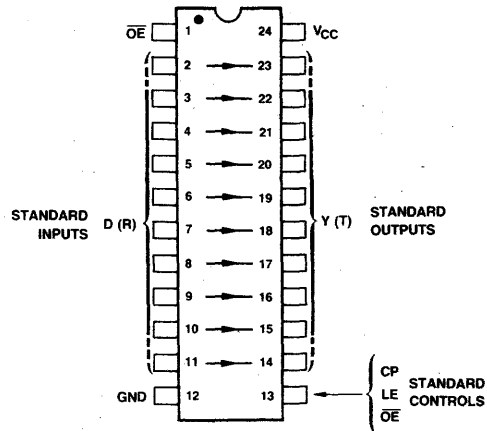
#### D-24-SLIM



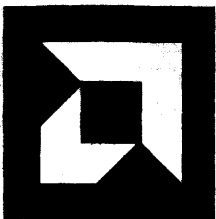
### HERMETIC DUAL IN-LINE PACKAGE

AMD Pkg	D-24-Slim	
Common Name	SLIM CERDIP	
38510 Appendix C	-	
Parameters	Min	Max
A	.140	.220
b	.016	.020
b <sub>1</sub>	.045	.065
c	.009	.011
D	1.230	1.285
E	.245	.285
E <sub>1</sub>	.290	.320
e	.090	.110
L	.120	.150
Q	.015	.060
S <sub>1</sub>	.010	
α	3°	13°

### STANDARDIZED PIN-OUTS FOR EASY BOARD LAYOUT



BLI-224



# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family

### Am29821/822/823/824/825/826

- High-speed parallel registers with positive edge-triggered D-type flip-flops
  - Noninverting CP-Y  $t_{PD} = 7.5\text{ns typ}$
  - Inverting CP-Y  $t_{PD} = 7.5\text{ns typ}$
- Buffered common Clock Enable ( $\overline{EN}$ )
- Buffered common asynchronous Clear input ( $\overline{CLR}$ )
- Three-state outputs glitch free during power-up and down
- Outputs have Schottky clamp to ground
- 48mA Commercial  $I_{OL}$ , 32mA MIL  $I_{OL}$
- High capacitance load capability
- Low capacitance inputs and outputs
- $I_{OH}$  specified 2.0V and 2.4V
- 24-pin 0.3" space saving package

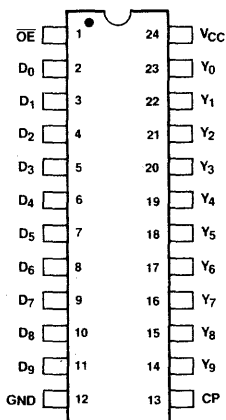
### FUNCTIONAL DESCRIPTION

The Am29820 Series bus interface registers are designed to eliminate the extra packages required to buffer existing registers and provide extra data width for wider address/data paths or buses carrying parity. The Am29821 and Am29822 are buffered, 10-bit wide version of the popular '374 function. The Am29823 and Am29824 are 9-bit wide buffered registers with Clock Enable ( $\overline{EN}$ ) and Clear ( $\overline{CLR}$ ) - ideal for parity bus interfacing in high performance microprogrammed systems. The Am29825 and Am29826 are 8-bit buffered registers with all the '823/4 controls plus multiple enables ( $\overline{OE}_1, \overline{OE}_2, \overline{OE}_3$ ) to allow multiuser control of the interface, e.g.,  $\overline{CS}$ , DMA, and RD/WR. They are ideal for use as an output port requiring high  $I_{OL}/I_{OH}$ .

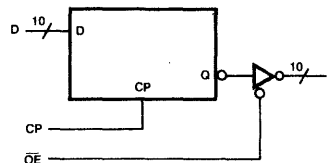
All of the Am29800 high performance interface family are designed for high capacitance load drive capability while providing low capacitance bus loading at both inputs and outputs. All inputs are Schottky diode inputs, and all outputs are designed for low capacitance bus loading in the high impedance state.

	Device		
	10-Bit	9-Bit	8-Bit
Noninverting	Am29821	Am29823	Am29825
Inverting	Am29822	Am29824	Am29826

### Am29821/Am29822 10-BIT REGISTERS

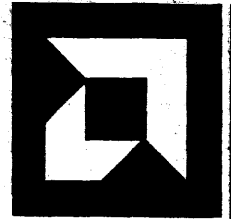


BLI-225



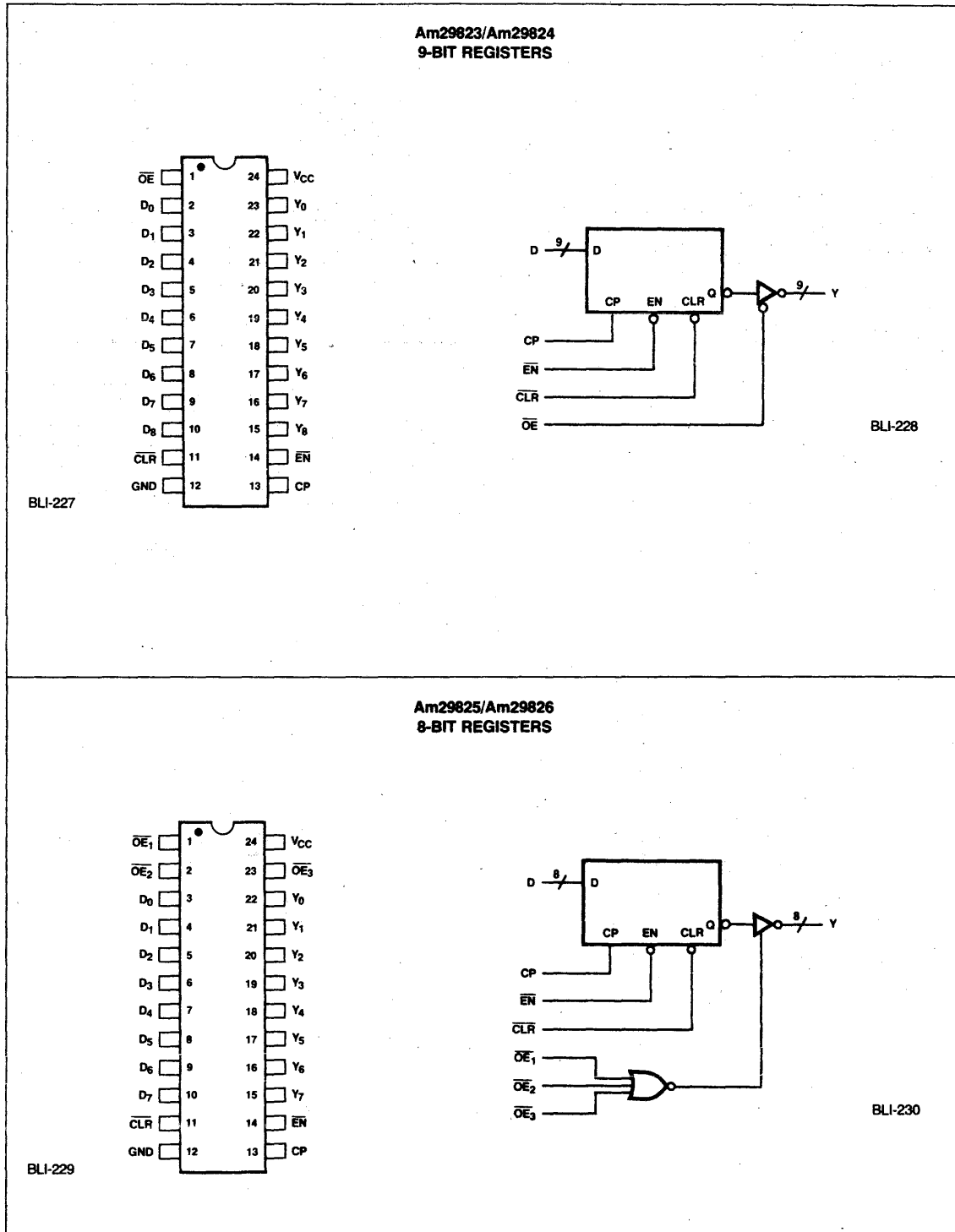
BLI-226

# Advanced Micro Devices



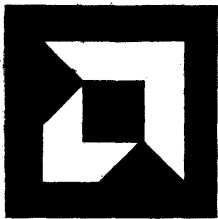
INTERFACE

## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family



Advanced Micro Devices





# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family

### Am29827/828, Am29861/862/863/864

- High-speed symmetrical bidirectional transceivers
  - Noninverting  $t_{PD} = 4.5ns$  typ
  - Inverting  $t_{PD} = 3.0ns$  typ
- High speed buffers and inverters
  - Noninverting  $t_{PD} = 4.5ns$  typ
  - Inverting  $t_{PD} = 3.0ns$  typ
- 200mV minimum input hysteresis on input data ports
- Three-state outputs glitch-free during power-up and down
- Outputs have Schottky clamp to ground
- 48mA Commercial  $I_{OL}$ , 32mA MIL  $I_{OL}$
- High capacitance load capability
- Low capacitance inputs and outputs
- $I_{OH}$  specified 2.0V and 2.4V
- 24-pin 0.3" space saving package

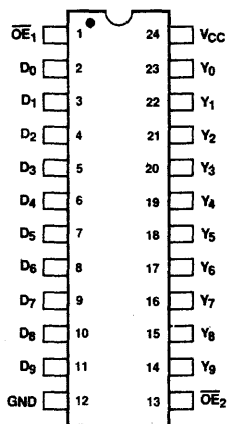
### FUNCTIONAL DESCRIPTION

The Am29827 and Am29828 10-bit bus buffers and Am29860 Series bus transceivers provide high performance bus interface buffering for wide data/address paths or buses carrying parity. The 10-bit buffers and 9-bit transceivers have NOR-ed output enables for maximum control flexibility. All buffer and transceiver data inputs have 200mV minimum input hysteresis to provide improved noise rejection.

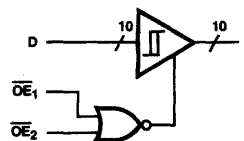
All of the Am29800 high performance interface family are designed for high capacitance load drive capability while providing low capacitance bus loading at both inputs and outputs. All inputs are Schottky diode inputs, and all outputs are designed for low capacitance bus loading in the high impedance state.

	Device		
	10-Bit Buffers	10-Bit Transceivers	9-Bit Transceivers
Noninverting	Am29827	Am29861	Am29863
Inverting	Am29828	Am29862	Am29864

### Am29827/Am29828 10-BIT BUS DRIVERS



BLI-237

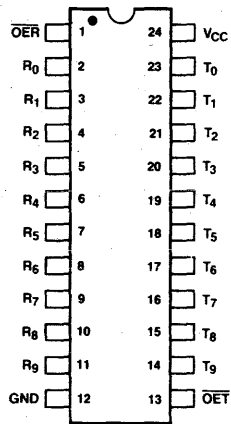


BLI-238

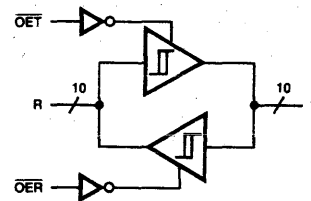


**BIPOLAR LSI AND SUPPORT PRODUCTS**  
**High Performance Bus Interface Circuits**  
**The Am29800 Family**

**Am29861/Am29862**  
**10-BIT TRANSCEIVERS**

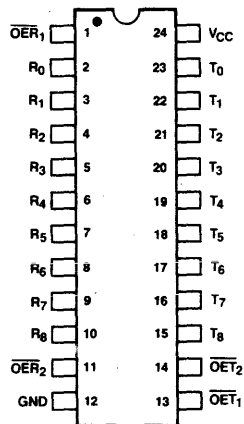


BLI-239

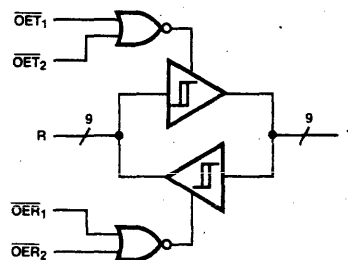


BLI-240

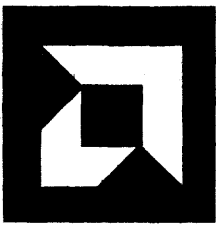
**Am29863/Am29864**  
**9-BIT TRANSCEIVERS**



BLI-241



BLI-242



# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family

### Am29833/834

- High speed bidirectional transceivers for 8-bit non-parity to 9-bit with parity ports
  - Noninverting data  $t_{PD} = 6.0ns$  typ
  - Inverting data  $t_{PD} = 6.0ns$  typ
  - Parity generate  $t_{PD} = 9.0ns$  typ
- High speed parity generation for Transmit mode
- High speed parity fault detection for Receive mode
- Clearable, open-collector output, Fault Flag register
- 200mV minimum input hysteresis
- Three-state outputs glitch-free during power-up and down
- Outputs have Schottky clamp to ground
- 48mA Commercial  $I_{OL}$ , 32mA MIL  $I_{OL}$
- High capacitance load capability
- Low capacitance inputs and outputs
- $I_{OH}$  specified 2.0V and 2.4V
- 24-pin 0.3" space saving package

### FUNCTIONAL DESCRIPTION

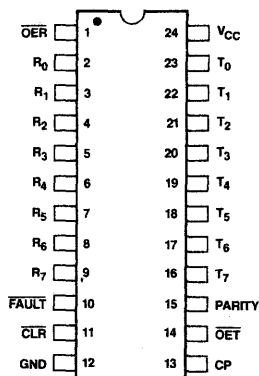
The Am29833 and Am29834 bidirectional transceivers are designed to interface an 8-bit data path without parity to a byte-parity 9-bit data path. All outputs -  $R_i$ ,  $T_i$  and PARITY - have high  $I_{OL}$  drive capability and are ideal for device-to-bus or bus-to-bus interfacing.

The internal  $\overline{FAULT}$  flag register is configured as a "one's catcher" to capture and hold any odd-parity fault occurring at the rising edge of the clock, CP. A registered  $\overline{FAULT}$  output remains LOW until cleared. Also, the  $\overline{FAULT}$  output is an open-collector output for wired-OR configurations where byte-parity is used for 16-bit or wider data buses or where multiple port flags are wired-OR tied together.

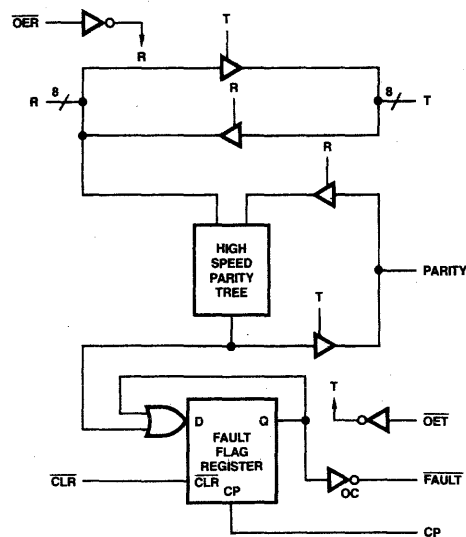
All of the Am29800 high performance interface family are designed for high capacitance load drive capability while providing low capacitance bus loading at both inputs and outputs. All inputs are Schottky diode inputs, and all outputs are designed for low capacitance bus loading in the high impedance state.

Device	
Noninverting	Am29833
Inverting	Am29834

### Am29833/Am29834 8-BIT TO 9-BIT PARITY TRANSCEIVERS

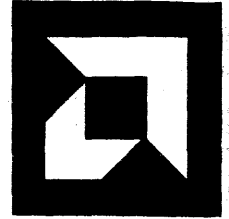


BLI-243



BLI-244

# Advanced Micro Devices



## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family

### Am29841/842/843/844/845/846

- High-speed parallel latches
  - Noninverting transparent  $t_{PD} = 4.5\text{ns typ}$
  - Inverting transparent  $t_{PD} = 6.0\text{ns typ}$
- Buffered common latch enable input
- Buffered common clear input
- Buffered common preset input
- Three-state outputs glitch-free during power-up and down
- Outputs have Schottky clamp to ground
- 48mA Commercial  $I_{OL}$ , 32mA MIL  $I_{OL}$
- High capacitance load capability
- Low capacitance inputs and outputs
- $I_{OH}$  specified 2.0V and 2.4V
- 24-pin 0.3" space saving package

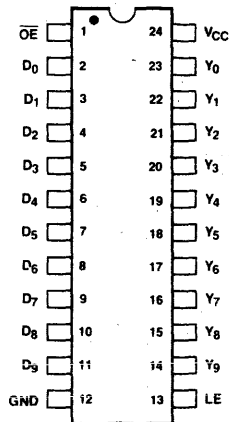
### FUNCTIONAL DESCRIPTION

The Am29840 Series bus interface latches are designed to eliminate the extra packages required to buffer existing latches and provide extra data width for wider address/data paths or buses carrying parity. The Am29841 and Am29842 are buffered, 10-bit wide version of the popular '373 function. The Am29843 and Am29844 are 9-bit wide buffered latches with Preset ( $\overline{PRE}$ ) and Clear ( $\overline{CLR}$ ) – ideal for parity bus interfacing in high performance systems. The Am29845 and Am29846 are 8-bit buffered latches with all the '843/4 controls plus multiple enables ( $\overline{OE}_1, \overline{OE}_2, \overline{OE}_3$ ) to allow multiuser control of the interface, e.g.,  $\overline{CS}$ , DMA, and RD/ $\overline{WR}$ . They are ideal for use as an output port requiring high  $I_{OL}/I_{OH}$ .

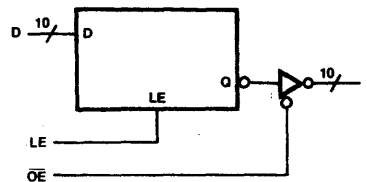
All of the Am29800 high performance interface family are designed for high capacitance load drive capability while providing low capacitance bus loading at both inputs and outputs. All inputs are Schottky diode inputs, and all outputs are designed for low capacitance bus loading in the high impedance state.

	Device		
	10-Bit	9-Bit	8-Bit
Noninverting	Am29841	Am29843	Am29845
Inverting	Am29842	Am29844	Am29846

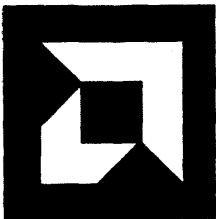
### Am29841/Am29842 10-BIT LATCHES



BLI-231



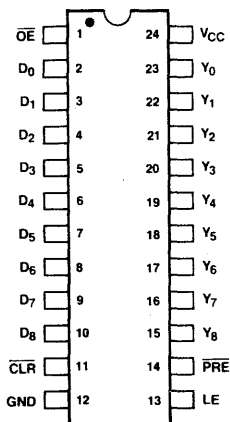
BLI-232



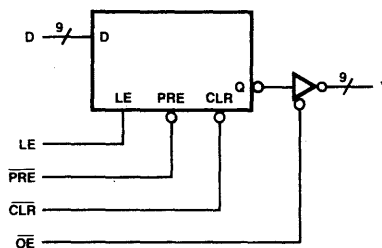
# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS High Performance Bus Interface Circuits The Am29800 Family

Am29843/Am29844  
9-BIT LATCHES

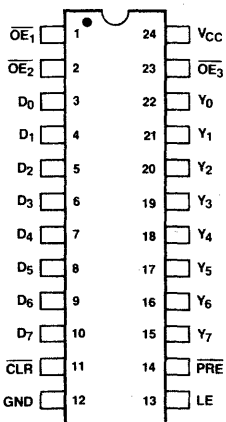


BLI-233

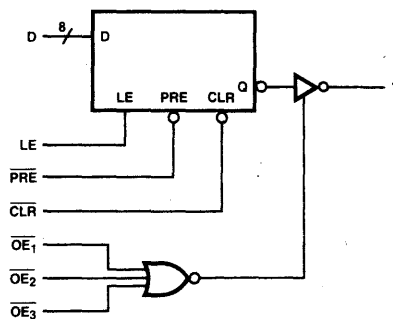


BLI-234

Am29845/Am29846  
8-BIT LATCHES



BLI-235



BLI-236

# ABBREVIATIONS OF COMPANY NAMES

<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.			<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Harris</b>	Harris Semiconductor	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Heurikon</b>	Heurikon Corp.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Hitachi</b>	Hitachi America, Ltd.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>Holt</b>	Holt Inc.	<b>Panasonic</b>	Panasonic
<b>Analogic</b>	Analogic	<b>HP</b>	Hewlett-Packard	<b>Pico Design</b>	Pico Design
<b>Analog Sys</b>	Analog Systems	<b>Hughes</b>	Hughes Aircraft, Solid State Products	<b>Polycore</b>	Polycore Electronics
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Plessey</b>	Plessey Semiconductors
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>PMI</b>	Precision Monolithics, Inc.
<b>APM</b>	Applied Microsystems Corp.			<b>PragDes</b>	Pragmatic Design Inc.
<b>Appl Sys</b>	Applied Systems Corp.	<b>ICC</b>	International Cybernetics	<b>Pro-Log</b>	Pro-Log Corp.
<b>APT</b>	Applied Microtechnology	<b>IDT</b>	Integrated Device Technology		
<b>Aptek</b>	Aptek Microsystems	<b>IMI</b>	Integrated Microcircuits, Inc.	<b>Quay</b>	Quay Corp.
<b>Array Tech</b>	Array Technology	<b>IMP</b>	International Microelectronic Products		
<b>AWI</b>	AWI Electronics	<b>IMS</b>	Industrial MicroSystems Inc.	<b>Raytheon</b>	Raytheon Semiconductor
		<b>Infosphere</b>	Infosphere	<b>RCA</b>	RCA Solid State Division
<b>Barvon</b>	Barvon Research	<b>Inmos</b>	Inmos	<b>RCI Data</b>	RCI Data
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>RELMS</b>	Relational Memory Systems
<b>Burr-Brown</b>	Burr-Brown	<b>IntCirSys</b>	Integrated Circuit Systems	<b>Reticon</b>	Reticon
		<b>IntCompSys</b>	Integrated Computer Systems	<b>RIFA</b>	RIFA
		<b>Int Tech</b>	Integrated Technology Corp.	<b>Rockwell</b>	Rockwell, Microelectronic Devices
<b>CAE</b>	Computer Aided Engineering	<b>Intech</b>	Intech Microcircuits Div.	<b>RTC</b>	Riehl Time Corporation
<b>Cal Devices</b>	California Devices	<b>Intel</b>	Intel		
<b>Cermetek</b>	Cermetek	<b>Interdesign</b>	Interdesign	<b>Sanyo</b>	Sanyo
<b>CGRS</b>	CGRS Microtech Inc.	<b>Intersil</b>	Intersil	<b>SBE</b>	SBE, Inc.
<b>Cherry</b>	Cherry Semiconductor	<b>Intronics</b>	Intronics	<b>SEEQ</b>	SEEQ Technology, Inc.
<b>CIC</b>	Custom Integrated Circuits	<b>ITT</b>	ITT Semiconductors	<b>SPI</b>	Semi Processes Inc.
<b>CirTech</b>	Circuit Technology			<b>Siemens</b>	Siemens
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<b>Comlinear</b>	Comlinear Corporation	<b>Kontron</b>	Kontron Electronics	<b>Signetics</b>	Signetics
<b>CMA</b>	Custom MOS Arrays			<b>SGS</b>	SGS Semiconductor
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<b>Data Trans</b>	Data Translation	<b>Micro Sci</b>	Micro Sciences Corp.	<b>Stynetic</b>	Stynetic Systems
<b>DateI</b>	DateI	<b>Microware</b>	Micro Sciences Corp.	<b>Sunrise</b>	Sunrise Electronics
<b>Datricon</b>	Datricon Corporation	<b>Microware</b>	Micro Sciences Corp.	<b>Sunshine</b>	Sunshine Semiconductor
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<b>ETI Micro</b>	ETI Micro	<b>Microware</b>	Micro Sciences Corp.	<b>TI</b>	Texas Instruments
<b>Exar</b>	Exar Integrated Systems	<b>Microware</b>	Micro Sciences Corp.	<b>Third Domain</b>	Third Domain
<b>Exel</b>	Exel Microelectronics	<b>Microware</b>	Micro Sciences Corp.	<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
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<b>Ferranti</b>	Ferranti Electric	<b>Microware</b>	Micro Sciences Corp.	<b>TRW</b>	TRW LSI Products
<b>Force</b>	Force Computers	<b>Microware</b>	Micro Sciences Corp.		
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		<b>Microware</b>	Micro Sciences Corp.		
<b>National</b>	National Semiconductor	<b>Microware</b>	Micro Sciences Corp.	<b>Varix</b>	Varix Corp.
<b>NCM</b>	NCM Corp.	<b>Microware</b>	Micro Sciences Corp.	<b>VLSI Design</b>	VLSI Design Associates
<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Microware</b>	Micro Sciences Corp.	<b>VTI</b>	VLSI Technology, Inc.
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<b>Nitron</b>	Nitron	<b>Microware</b>	Micro Sciences Corp.		
		<b>Microware</b>	Micro Sciences Corp.	<b>Weitek</b>	Weitek Corporation
		<b>Microware</b>	Micro Sciences Corp.	<b>Western</b>	Western Digital
		<b>Microware</b>	Micro Sciences Corp.	<b>Wintek</b>	Wintek Corp.
		<b>Microware</b>	Micro Sciences Corp.		
		<b>Microware</b>	Micro Sciences Corp.	<b>Xicom</b>	Xicom, Inc.
		<b>Microware</b>	Micro Sciences Corp.	<b>Xycom</b>	Xycom
		<b>Microware</b>	Micro Sciences Corp.		
		<b>Microware</b>	Micro Sciences Corp.	<b>Zendex</b>	Zendex Corp.
		<b>Microware</b>	Micro Sciences Corp.	<b>Zilog</b>	Zilog
		<b>Microware</b>	Micro Sciences Corp.	<b>ZyMOS</b>	ZyMOS Corporation
		<b>Microware</b>	Micro Sciences Corp.	<b>Zytrex</b>	Zytrex Corp.



## 32 BIT DRIVER

### Features

- Drives Up to 32 Devices
- Cascadable
- On Chip Oscillator
- Requires Only 3 Control Lines
- CMOS Construction For:
  - Wide Supply Range
  - High Noise Immunity
  - Wide Temperature Range

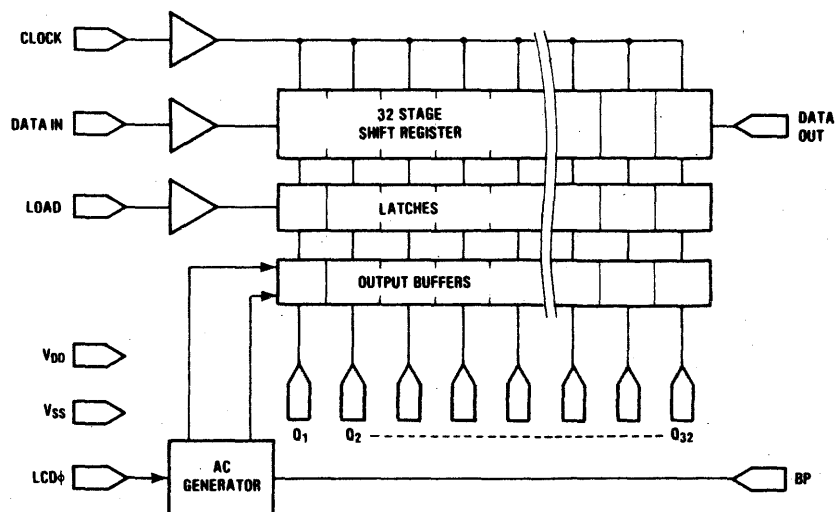
### Applications:

- Liquid Crystal Displays
- LED and Incandescent Displays
- Solenoids
- Print Head Drives
- DC and Stepping Motors
- Relays

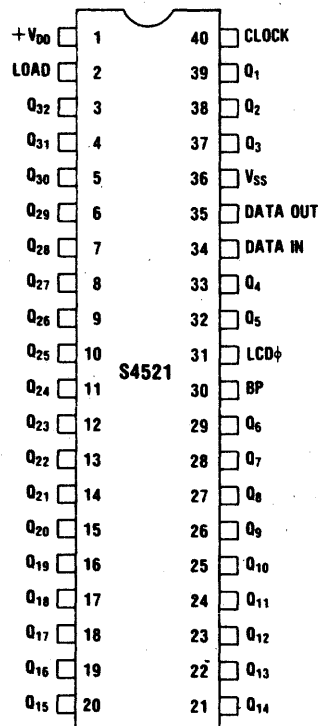
### General Description

The AMI S4521 is an MOS/LSI circuit that drives a variety of output devices, usually under microprocessor control. This device requires only three control lines due to its serial input construction. It latches the data to be output, relieving the microprocessor from the task of generating the required waveform, or it may be used to bring data directly to the drivers. The part acts as a versatile peripheral to drive displays, motors, relays and solenoids within its output limitations. It is especially well suited to driving liquid crystal displays, with a backplane A.C. signal option that is provided. The A.C. frequency of the backplane output that can be user supplied or generated by attaching a capacitor to the LCD $\phi$  input, which controls the frequency of the internal oscillator. One circuit will drive up to 32 devices and more can be driven by cascading several drivers together.

### Functional Block Diagram



### Pin Configuration





## 10 BIT, HIGH VOLTAGE HIGH CURRENT DRIVER

### Features

- Outputs Capable of 60 Volt Swings at 25mA
- Drives Up to 10 Devices
- Cascadable
- Requires Only 4 Control Lines

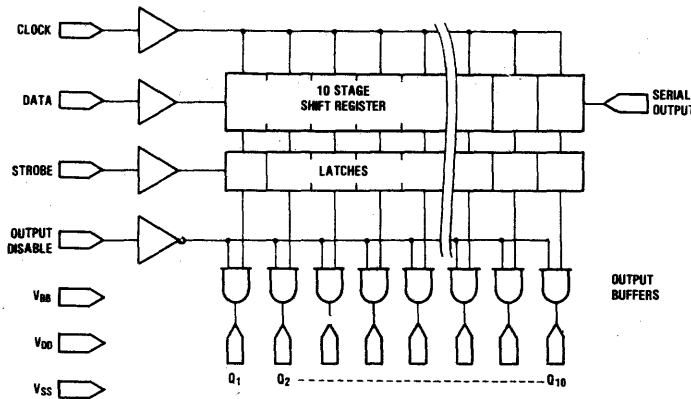
### Applications:

- Vacuum Fluorescent Displays
- LED and Incandescent Displays
- Solenoids
- Print Head Drives
- DC and Stepping Motors
- Relays

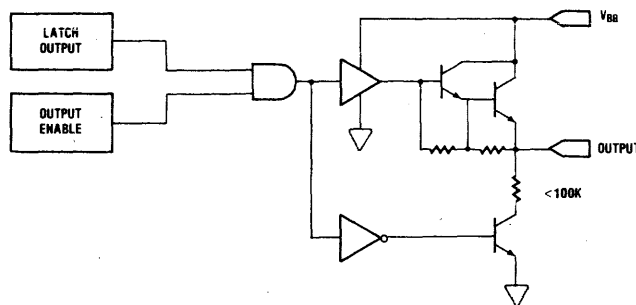
### General Description

The AMI S4534 is a high voltage, high current MOS/LSI circuit that drives a variety of output devices, usually under microprocessor control, by converting low level signals such as TTL, and CMOS to high voltage, high current drive signals. This device requires only four control lines due to its serial input construction. It latches the data to be output, or it may be used to bring data directly to the driver. The part acts as a versatile peripheral to drive displays, motors, relays and solenoids within its output limitations of a 60 volt swing and up to 50mA per drive. It is especially well suited to drive vacuum fluorescent displays due to its high voltage output capability. One circuit will drive up to 10 devices and more can be driven by cascading several drivers together.

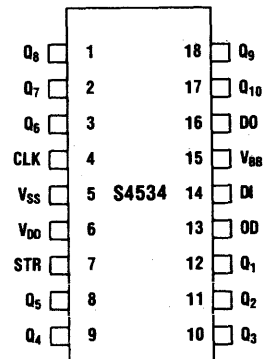
### Functional Block Diagram



### Output Buffer (Functional Diagram)



### Pin Configuration







## 32 BIT, HIGH VOLTAGE DRIVER

### Features

- High Voltage Outputs Capable of 60 Volt Swing
- Drives Up to 32 Devices
- Cascadable
- Requires Only 4 Control Lines

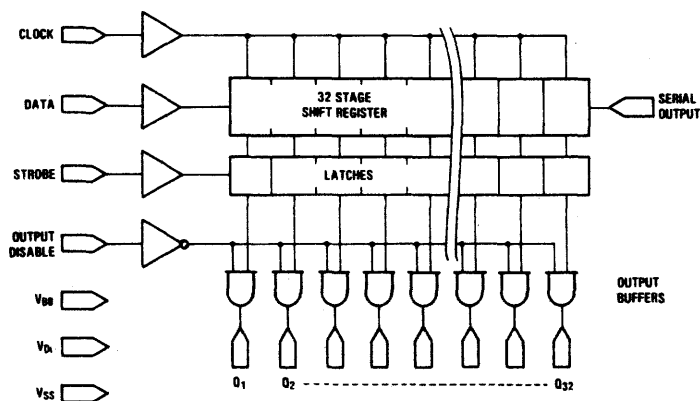
### Applications:

- Vacuum Fluorescent Displays
- LED and Incandescent Displays
- Solenoids
- Print Head Drives
- DC and Stepping Motors
- Relays

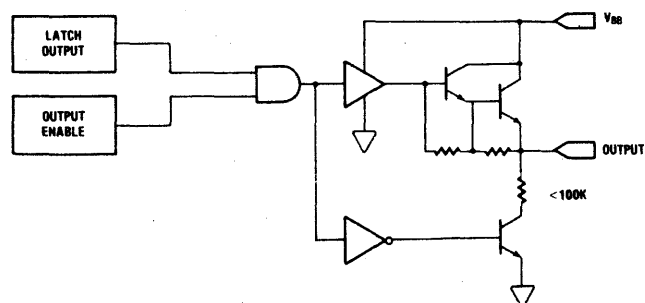
### General Description

The AMI S4535 is a high voltage MOS/LSI circuit that drives a variety of output devices, usually under microprocessor control, by converting low level signals such as TTL, and CMOS to high voltage, high current drive signals. This device requires only four control lines due to its serial input construction. It latches the data to be output, or it may be used to bring data directly to the driver. The part acts as a versatile peripheral to drive displays, motors, relays and solenoids within its output limitations of a 60 volt swing and up to 25mA per drive. It is especially well suited to drive vacuum fluorescent displays due to its high voltage output capability. One circuit will drive up to 32 devices and more can be driven by cascading several drivers together.

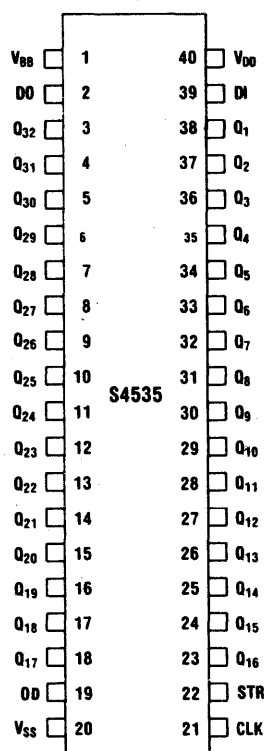
### Functional Block Diagram

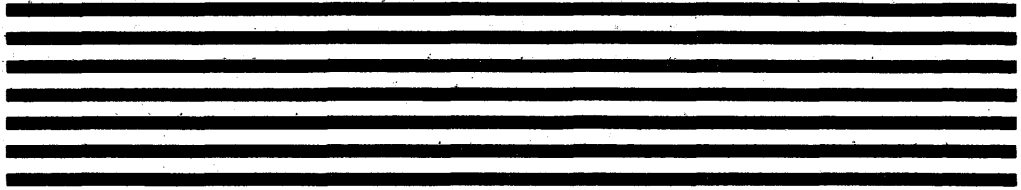


### Output Buffer (Functional Diagram)



### Pin Configuration





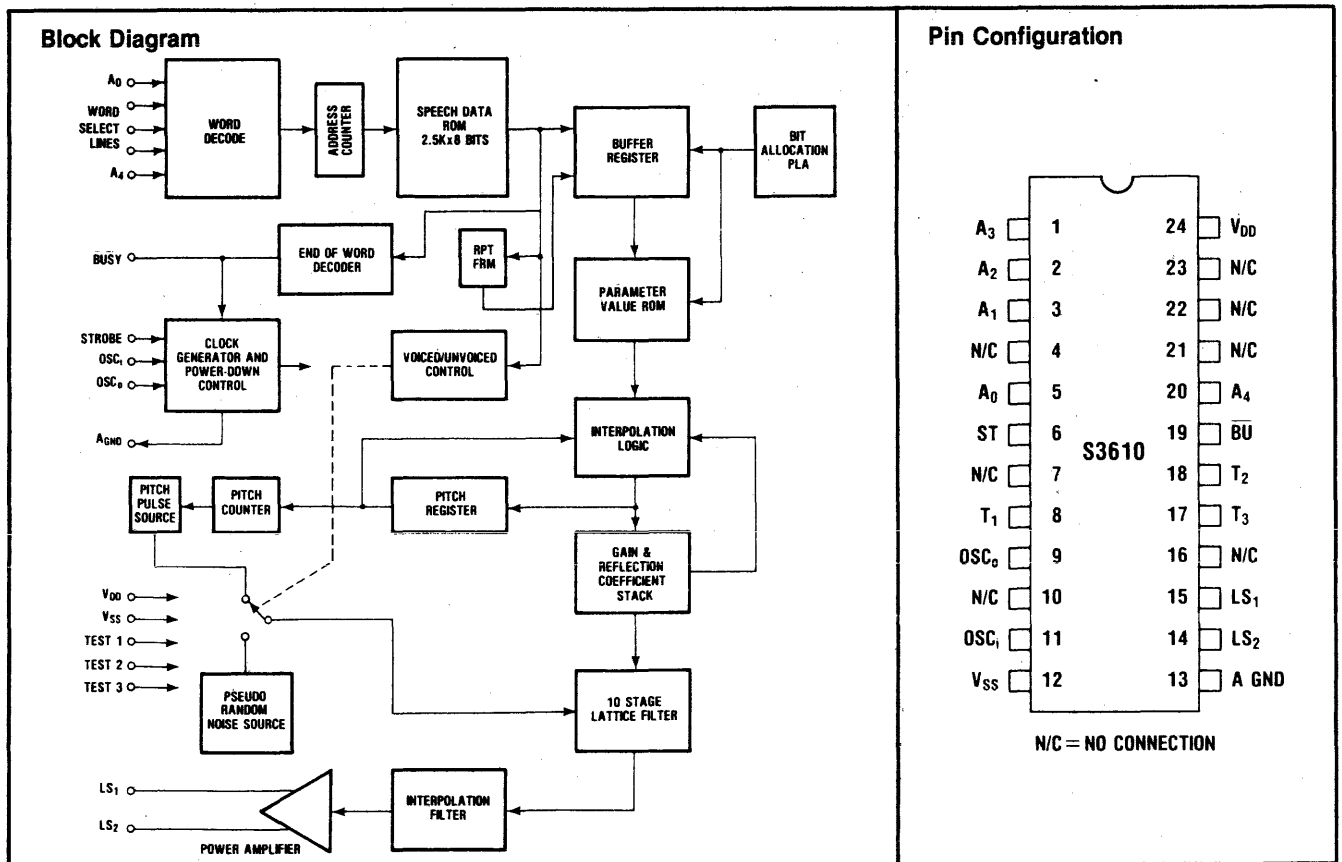
## LPC-10 SPEECH SYNTHESIZER WITH ON-CHIP 20K SPEECH DATA ROM

### Features

- Simple Digital Interface
- CMOS Switched-Capacitor Filter Technology
- Automatic Powerdown
- 5-8 Volts Single Power Supply Operation
- Direct Loudspeaker Drive
- 20mW Audio Output
- 20K Bits Speech ROM
- Low Data Rate
- Up to 32 Word Vocabulary

### General Description

The S3610 LPC-10 Speech Synthesizer generates speech of high quality and intelligibility from LPC (Linear Predictive Coding) data stored in an internal 20K bit ROM. The simple digital interface consists of 5 word-select lines, a strobe input to load the address data and initiate operation, and a busy output signal. At the end of enunciation the chip automatically goes into the powerdown mode until a new word select address is strobed in. The data rate from the speech ROM into the synthesizer is 2.0K bits/sec max. Typically the average data rate will be reduced to about 1.2K bits/sec. by means of the data rate reduction techniques used internally.





## LPC-10 SPEECH SYNTHESIZER

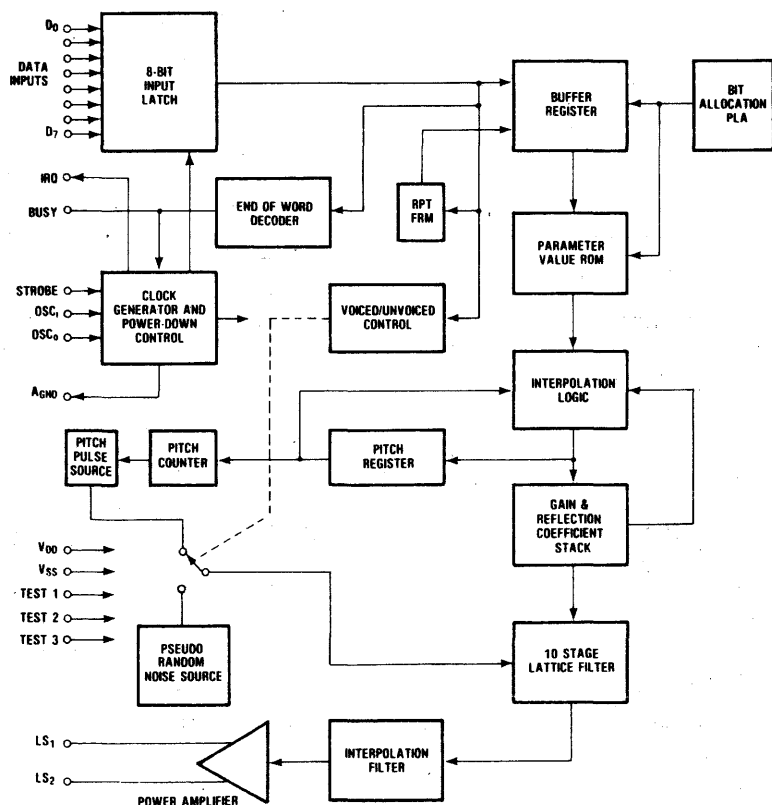
### Features

- Simple Microprocessor Interface
- CMOS Switched-Capacitor Filter Technology
- Automatic Powerdown
- 5-8 Volts Single Power Supply Operation
- Direct Loudspeaker Drive
- 20mW Audio Output
- Low Data Rate

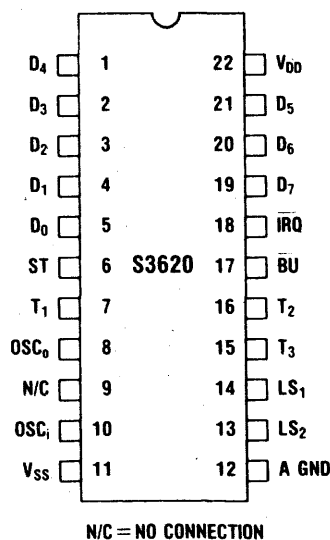
### General Description

The S3620 LPC-10 Speech Synthesizer generates speech of high quality and intelligibility from LPC (Linear Predictive Coding) data stored in an external memory. The digital interface circuitry is fully microprocessor compatible and allows the processor to load the data with or without a DMA controller. The loading takes place on a handshake basis, and in the absence of a response from the processor the synthesizer automatically shuts down and goes into the powerdown mode. A busy signal allows the processor to sense the status of the synthesizer.

### Block Diagram



### Pin Configuration



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HEARST BUSINESS COMMUNICATIONS, INC./UTP DIVISION

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# Quad 12-Bit Microprocessor-Compatible D/A Converter

## AD390\*

### FEATURES

- Four Complete 12-Bit DACs in One IC Package
- Linearity Error  $\pm 1/2$ LSB  $T_{min} - T_{max}$  (AD390K, T)
- Factory-Trimmed Gain and Offset
- Buffered Voltage Output
- Monotonicity Guaranteed Over Full Temperature Range
- Double-Buffered Data Latches
- Includes Reference and Buffer
- Fast Settling: 8 $\mu$ s max to  $\pm 1/2$ LSB
- Low Cost

### PRODUCT DESCRIPTION

The AD390 contains four 12-bit high speed voltage-output digital-to-analog converters in a compact 28-pin hybrid package. The design is based on a proprietary latched 12-bit DAC chip which reduces chip count and provides high reliability. The AD390 is ideal for systems requiring digital control of many analog voltages where board space is at a premium. Such applications include automatic test equipment, process controllers, and vector-scan displays.

The AD390 is laser-trimmed to  $\pm 1/2$ LSB max nonlinearity (AD390KD, TD) and absolute accuracy of  $\pm 0.05$  percent of full scale. The high initial accuracy is made possible by the use of thin-film scaling resistors on the monolithic DAC chips. The internal buried zener voltage reference provides excellent temperature drift characteristics (20ppm/ $^{\circ}$ C) and an initial tolerance of  $\pm 0.03\%$  maximum. The internal reference buffer allows a single common reference to be used for multiple AD390 devices in large systems.

The individual DACs are accessed by the  $\overline{CS1}$  through  $\overline{CS4}$  control inputs and the  $\overline{A0}$  and  $\overline{A1}$  lines. These control signals permit the registers of the four DACs to be loaded sequentially and the outputs to be simultaneously updated.

The AD390 outputs are calibrated for a  $\pm 10$ V output range with positive-true offset binary input coding. A 0 to +10V version is available on special order.

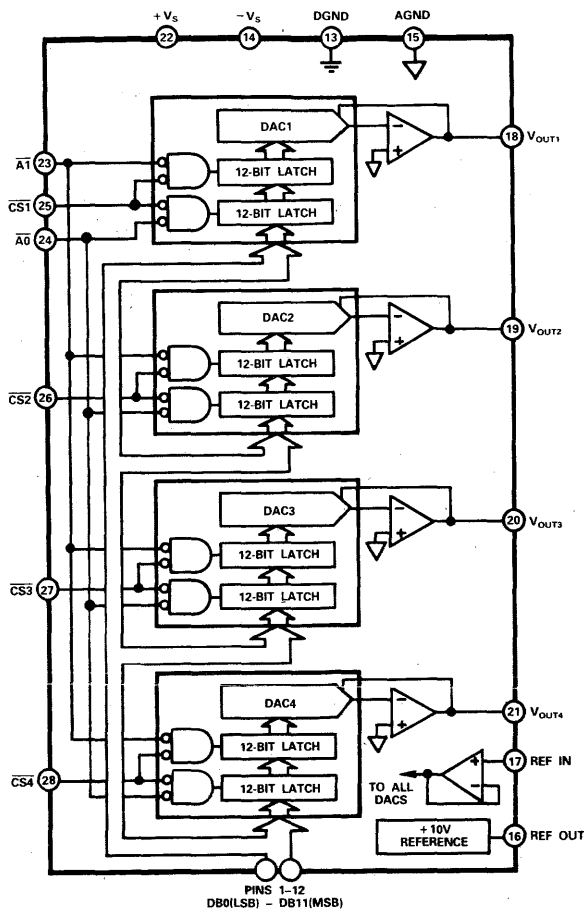
The AD390 is packaged in a 28-lead ceramic package and is specified for operation over the 0 to +70 $^{\circ}$ C and -55 $^{\circ}$ C to +125 $^{\circ}$ C temperature range.

### ORDERING GUIDE

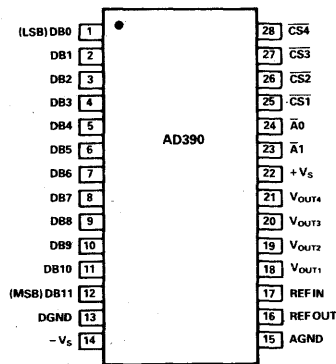
Model	Temperature Range	Gain Error	Linearity Error
		25 $^{\circ}$	$T_{min} - T_{max}$
AD390JD	0 to +70 $^{\circ}$ C	$\pm 4$ LSB	$\pm 3/4$ LSB
AD390KD	0 to +70 $^{\circ}$ C	$\pm 2$ LSB	$\pm 1/2$ LSB
AD390SD	-55 $^{\circ}$ C to +125 $^{\circ}$ C	$\pm 4$ LSB	$\pm 3/4$ LSB
AD390TD	-55 $^{\circ}$ C to +125 $^{\circ}$ C	$\pm 2$ LSB	$\pm 1/2$ LSB

\*Covered by patent numbers 3,803,590; 3,890,611; 3,932,863; 3,978,473; 4,020,486 and other patents pending.

### AD390 FUNCTIONAL BLOCK DIAGRAM



### PIN CONFIGURATION TOP VIEW



#### FEATURES

- Complete 8-Bit DAC
- Voltage Output – 2 Calibrated Ranges
- Internal Precision Band-Gap Reference
- Single-Supply Operation: +5V to +15V
- Full Microprocessor Interface
- Fast: 1μs Voltage Settling to ±1/2LSB
- Low Power: 75mW
- No User Trims
- Guaranteed Monotonic Over Temperature
- All Errors Specified T<sub>min</sub> to T<sub>max</sub>
- Small 16-Pin DIP Package
- Single Laser-Wafer-Trimmed Chip for Hybrids
- Low Cost

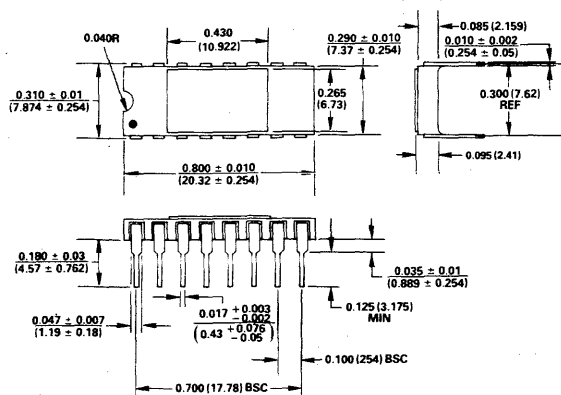
#### PRODUCT DESCRIPTION

The AD558 DACPORT is a complete voltage-output 8-bit digital-to-analog converter, including output amplifier, full microprocessor interface and precision voltage reference on a single monolithic chip. No external components or trims are required to interface, with full accuracy, an 8-bit data bus to an analog system.

The performance and versatility of the DACPORT is a result of several recently-developed monolithic bipolar technologies. The complete microprocessor interface and control logic is implemented with integrated injection logic (I<sup>2</sup>L), an extremely dense and low-power logic structure that is process-compatible with linear bipolar fabrication. The internal precision voltage reference is the patented low-voltage band-gap circuit which permits full-accuracy performance on a single +5V to +15V power supply. Thin-film silicon-chromium resistors provide the stability required for guaranteed monotonic operation over the entire operating temperature range (all grades), while recent advances in laser-wafer-trimming of these thin-film resistors permit absolute calibration at the factory to within ±1LSB; thus no user-trims for gain or offset are required. A new circuit design provides voltage settling to ±1/2LSB for a full-scale step in 800ns.

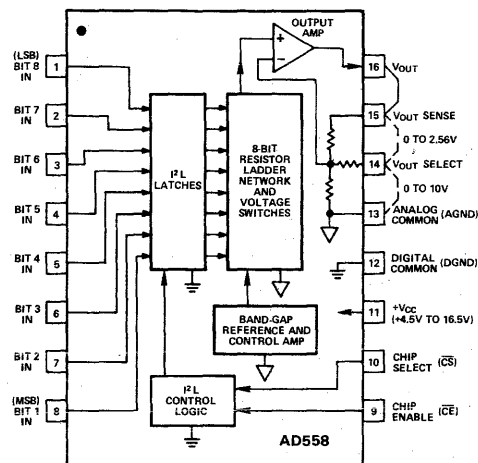
#### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



\*Covered by U.S. Patent Nos. 3,887,863; 3,685,045; 4,323,795;  
Patents Pending. DACPORT is a trademark of Analog Devices, Inc.

#### AD558 FUNCTIONAL BLOCK DIAGRAM



#### PRODUCT HIGHLIGHTS

1. The 8-bit I<sup>2</sup>L input register and fully microprocessor-compatible control logic allow the AD558 to be directly connected to 8- or 16-bit data buses and operated with standard control signals. The latch may be disabled for direct DAC interfacing.
2. The laser-trimmed on-chip SiCr thin-film resistors are calibrated for absolute accuracy and linearity at the factory. Therefore, no user trims are necessary for full rated accuracy over the operating temperature range.
3. The inclusion of a precision low-voltage band-gap reference eliminates the need to specify and apply a separate reference source.
4. The voltage-switching structure of the AD558 DAC section along with a high-speed output amplifier and laser-trimmed resistors give the user a choice of 0V to +2.56V or 0V to +10V output ranges, selectable by pin-strapping. Circuitry is internally compensated for minimum settling time on both ranges; typically settling to ±1/2LSB for a full-scale 2.55 volt step in 800ns.
5. The AD558 is designed and specified to operate from single +4.5V to +16.5V power supply.

#### AD558 ORDERING GUIDE

Model	Temperature	Relative Accuracy Error Max T <sub>min</sub> to T <sub>max</sub>	Full-Scale Error, Max T <sub>min</sub> to T <sub>max</sub>
AD558JN	0 to +70°C	± 1/2LSB	± 2.5LSB
AD558KN	0 to +70°C	± 1/4LSB	± 1LSB
AD558JD	0 to +70°C	± 1/2LSB	± 2.5LSB
AD558KD	0 to +70°C	± 1/4LSB	± 1LSB
AD558SD	-55°C to +125°C	± 3/4LSB	± 2.5LSB
AD558TD	-55°C to +125°C	± 3/8LSB	± 1LSB

### FEATURES

- Single Chip Construction
- Double-Buffered Latch for 8-Bit  $\mu$ P-Compatibility
- On Chip Output Amplifier
- High Stability Buried Zener Reference On Chip
- Monotonicity Guaranteed Over Temperature
- Linearity Guaranteed Over Temperature: 1/2LSB max (AD667K)
- Guaranteed for Operation with  $\pm 12V$  or  $\pm 15V$  Supplies
- Low Power
- TTL/5V CMOS Compatible Logic Inputs
- Low Cost

### PRODUCT DESCRIPTION

The AD667 is a complete voltage output 12-bit digital-to-analog converter including a high stability buried zener voltage reference, an output amplifier and double-buffered input latch on a single chip. The converter uses 12 precision high speed bipolar current steering switches and a laser trimmed thin film resistor network to provide fast settling time and high accuracy.

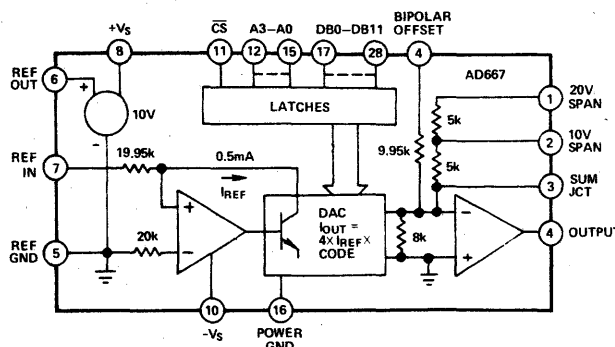
Microprocessor compatibility is achieved by the on-chip double-buffered latch. The design of the input latch allows direct interface to 4-, 8-, 12-, or 16-bit buses. The 12 bits of data from the first rank of latches can then be transferred to the second rank, avoiding generation of spurious analog output values. The latch responds to strobe pulses as short as 100ns, allowing use with the fastest available microprocessors.

The functional completeness and high performance in the AD667 results from a combination of advanced switch design, high speed bipolar manufacturing process, and the proven laser wafer-trimming (LWT) technology. The AD667 is trimmed at the wafer level and is specified to  $\pm 1/4$ LSB maximum linearity error (K grade) at 25°C and  $\pm 1/2$ LSB over the full operating temperature range.

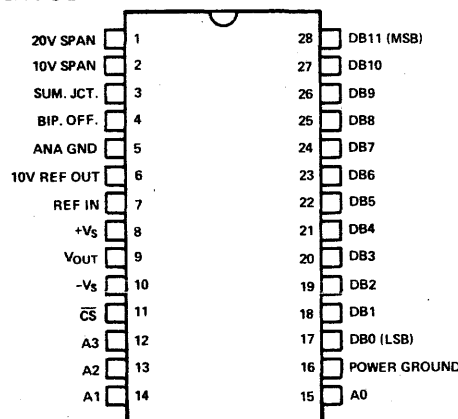
The subsurface (buried) Zener diode on the chip provides a low-noise voltage reference which has long-term stability and temperature drift characteristics comparable to the best discrete reference diodes. The laser trimming process which provides the excellent linearity is also used to trim both the absolute value of the reference as well as its temperature coefficient. The AD667 is thus well suited for wide temperature range performance with  $\pm 1/2$ LSB maximum linearity error and guaranteed monotonicity over the full temperature range. Typical full scale gain T.C. is 10ppm/°C.

The AD667 is available in four performance grades. The AD667J and K are specified for use over the 0 to +70°C temperature range and is available in a 28-pin hermetically-sealed, ceramic DIP. The AD667S grade is specified for the -55°C to +125°C range and is available in the ceramic package.

### AD667 FUNCTIONAL BLOCK DIAGRAM



### AD667 PINOUT

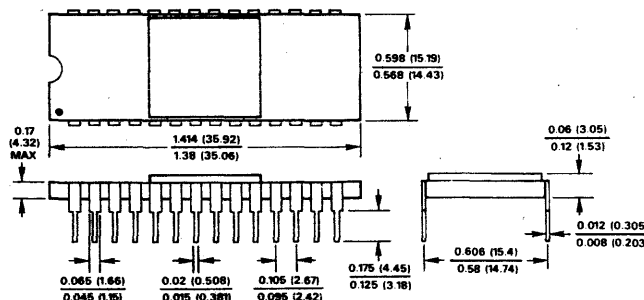


### AD667 ORDERING GUIDE

Model	Package	Temp. Range	Linearity Error max @ 25°C	Gain T.C. max
AD667JD	Ceramic	Com	$\pm 1/2$ LSB	50ppm/°C
AD667KD	Ceramic	Com	$\pm 1/4$ LSB	20ppm/°C
AD667SD	Ceramic	Mil	$\pm 1/2$ LSB	30ppm/°C

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



**FEATURES**

- On-Chip Latches for Both DACs
- +5V to +15V Operation
- DACs Matched to 1%
- Four Quadrant Multiplication
- TTL/CMOS Compatible
- Latch Free (Protection Schottkys not Required)

**APPLICATIONS**

- Digital Control of:
  - Gain/Attenuation
  - Filter Parameters
  - Stereo Audio Circuits
  - X-Y Graphics

**GENERAL DESCRIPTION**

The AD7528 is a monolithic dual 8-bit digital/analog converter produced in a small 0.3" wide 20-pin DIP, featuring excellent DAC-to-DAC matching.

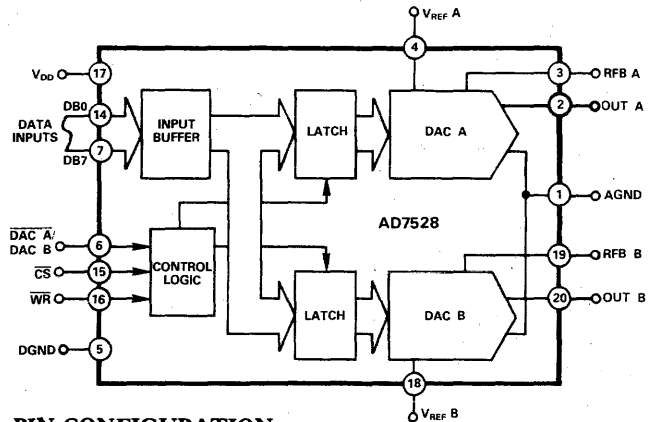
Separate on-chip latches are provided for each DAC to allow easy microprocessor interface.

Data is transferred into either of the two DAC data latches via a common 8-bit TTL/CMOS compatible input port. Control input DAC A/DAC B determines which DAC is to be loaded. The AD7528's load cycle is similar to the write cycle of a random access memory and the device is bus compatible with most 8-bit microprocessors, including 6800, 8080, 8085, Z80.

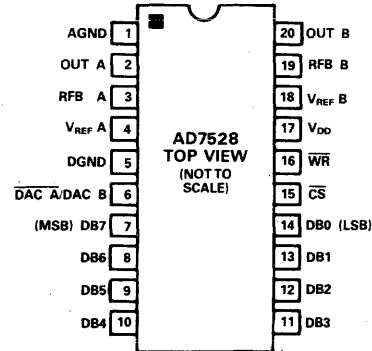
The device operates from a +5V to +15V power supply, dissipating only 20mW of power.

Both DACs offer excellent four quadrant multiplication characteristics with a separate reference input and feedback resistor for each DAC.

**AD7528 FUNCTIONAL BLOCK DIAGRAM**



**PIN CONFIGURATION**



**ORDERING INFORMATION**

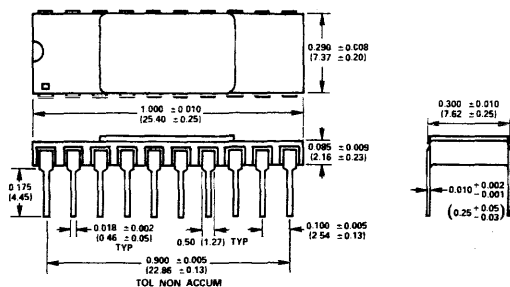
Relative Accuracy	Gain Error T <sub>A</sub> = +25°C	Temperature Range and Package		
		Plastic 0 to +70°C	Cerdip -25°C to +85°C	Ceramic -55°C to +125°C
± 1LSB	± 4LSB	AD7528JN	AD7528AQ	AD7528SD
± 1/2LSB	± 2LSB	AD7528KN	AD7528BQ	AD7528TD
± 1/2LSB	± 1LSB	AD7528LN	AD7528CQ	AD7528UD

**MECHANICAL INFORMATION**

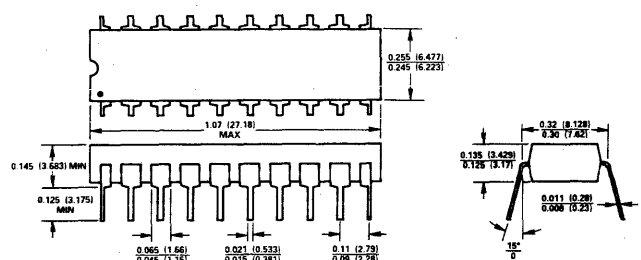
**OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).

**20 PIN CERAMIC (SUFFIX D)**



**20 PIN PLASTIC DIP (SUFFIX N)**







# LC<sup>2</sup>MOS High Speed 12-Bit Voltage DAC

## AD7240

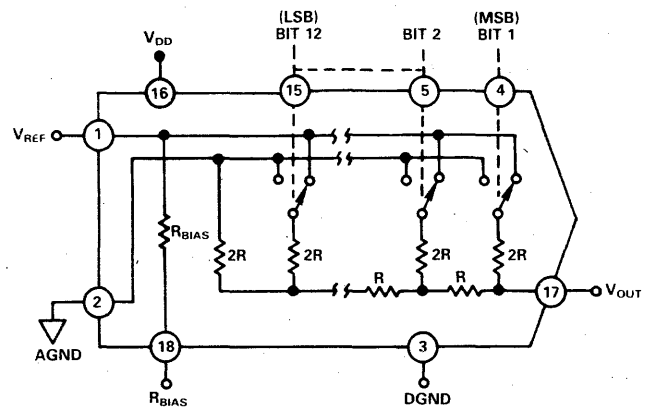
### FEATURES

- Fast Voltage Settling Time: 550ns to 0.01%
- Total Unadjusted Error: 1LSB max
- Single Supply Operation
- Latch Up Proof (No Protection Schottky Required)
- Superb Differential Nonlinearity: 1/2LSB max over Temperature
- Low Power Dissipation: 30mW

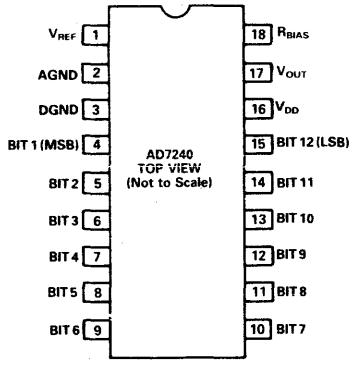
### APPLICATIONS

- Battery Powered Instrumentation
- High Speed A/D Converters
- Programmable Gain Amplifiers
- Vector Graphics
- S/D Converters

### AD7240 FUNCTIONAL BLOCK DIAGRAM



### PIN CONFIGURATION



### GENERAL DESCRIPTION

The Analog Devices AD7240 is a fast settling (550ns typically to 1/2LSB) 12-bit voltage output digital to analog converter. It is fabricated using an advanced high speed Linear Compatible CMOS process (LC<sup>2</sup>MOS) which has been specifically developed to allow high speed digital logic circuits and precision analog circuits to be integrated on the same chip.

The AD7240 operates with single +15 volts V<sub>DD</sub> supply and exhibits exceptionally fast settling times due to the small (and code independent) value of capacitance at the output of the DAC.

The AD7240 also gives superior performance to other CMOS DACs when configured in the current steering mode as a multi-plying DAC.

### ORDERING INFORMATION

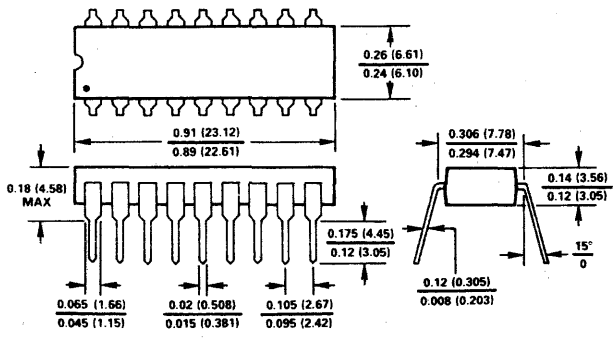
Total Unadjusted Error	Plastic	Cerdip	Side Brazed Ceramic
T <sub>A</sub> = T <sub>MIN</sub> to T <sub>MAX</sub>	0 to +70°C	-25°C to +85°C	-55°C to +125°C
+1/2, -2LSB	AD7240JN	AD7240AQ	AD7240SD
+1/4, -1 1/2LSB	AD7240KN	AD7240BQ	AD7240TD

### MECHANICAL INFORMATION

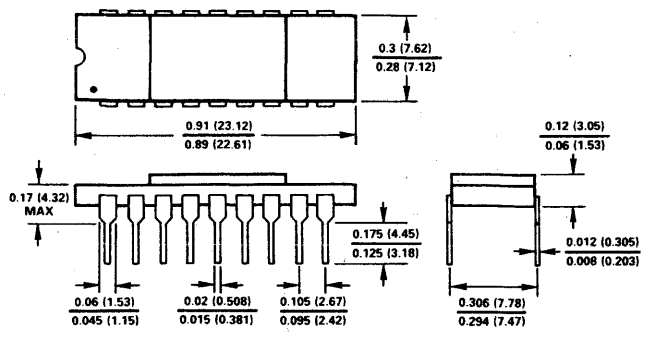
#### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

#### 18-PIN PLASTIC DIP - SUFFIX N



#### 18-PIN CERAMIC DIP - SUFFIX D



**FEATURES**

- 12-Bit Resolution
- Low Gain T.C.: 2ppm/°C typ
- Fast TTL Compatible Data Latches
- Single +5V to +15V Supply
- Small 20-Pin 0.3" DIP
- Latch Free (Schottky Protection Diode Not Required)
- Low Cost
- Ideal for Battery Operated Equipment

**GENERAL DESCRIPTION**

The AD7545 is a monolithic 12-bit CMOS multiplying DAC with on-board data latches. It is loaded by a single 12-bit wide word and interfaces directly to most 12- and 16-bit bus systems. Data is loaded into the input latches under the control of the CS and WR inputs; tying these control inputs low makes the input latches transparent allowing direct unbuffered operation of the DAC.

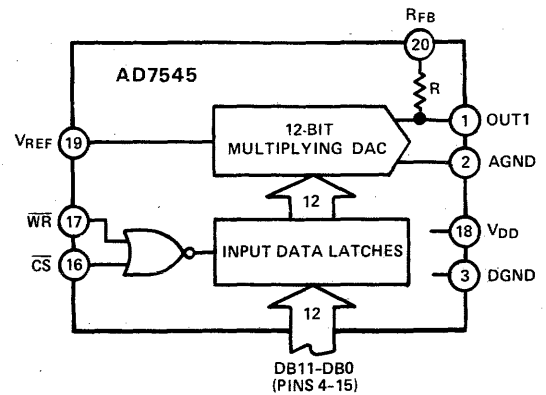
The AD7545 is particularly suitable for single supply operation and applications with wide temperature variations.

The AD7545 can be used with any supply voltage from +5V to +15V. With CMOS logic levels at the inputs the device dissipates less than 0.5mW for  $V_{DD} = +5V$ .

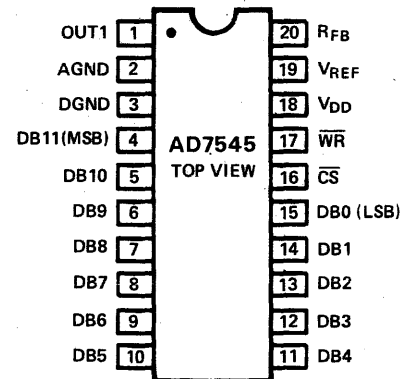
**ORDERING INFORMATION**

Relative Accuracy	Maximum Gain Error $T_A = +25^\circ C$ $V_{DD} = +5V$	Temperature Range and Package		
		Plastic 0 to +70°C	Cerdip -25°C to +85°C	Ceramic -55°C to +125°C
±2LSB	±20LSB	AD7545JN	AD7545AQ	AD7545SD
±1LSB	±10LSB	AD7545KN	AD7545BQ	AD7545TD
±1/2LSB	±5LSB	AD7545LN	AD7545CQ	AD7545UD
±1/2LSB	±1LSB	AD7545GLN	AD7545GCQ	AD7545GUD

**AD7545 FUNCTIONAL BLOCK DIAGRAM**



**PIN CONFIGURATION**

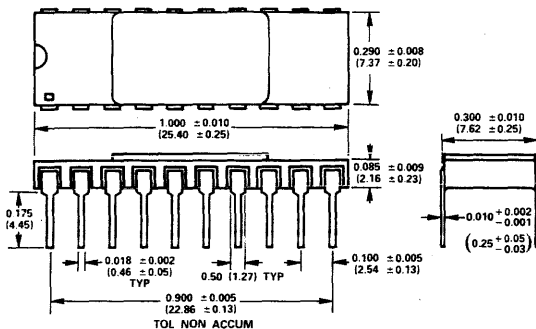


**MECHANICAL INFORMATION**

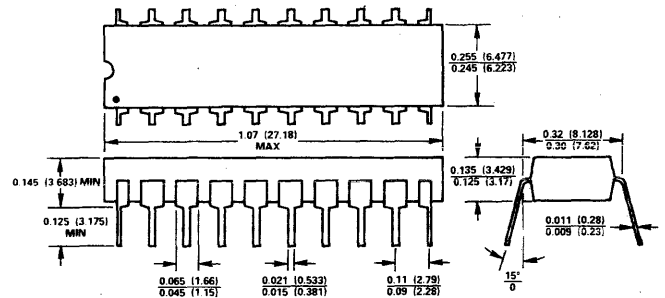
**OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).

**20-PIN CERAMIC DIP (SUFFIX D)**



**20-PIN PLASTIC DIP (SUFFIX N)**



### FEATURES

- 8-Bit Bus Compatible 12-Bit DAC
- All Grades 12-Bit Monotonic Over Full Temperature Ranges
- Operation Specified at +5V, +12V or +15V Power Supply
- Low Gain Drift of 5ppm/ $^{\circ}$ C Maximum
- Full 4 Quadrant Multiplication
- Small 20-Pin Package

### APPLICATIONS

- 8-Bit Microprocessor Based Control Systems
- Programmable Amplifiers
- Function Generation
- Servo Control

### GENERAL DESCRIPTION

The AD7548 is a 12-bit monolithic CMOS D/A converter for use with 8-bit bus microprocessors. Data is loaded in two bytes to input holding registers as shown in the block diagram opposite. The AD7548 can be configured to accept either left- or right-justified data, least significant byte or most significant byte first, using standard TTL compatible control inputs.

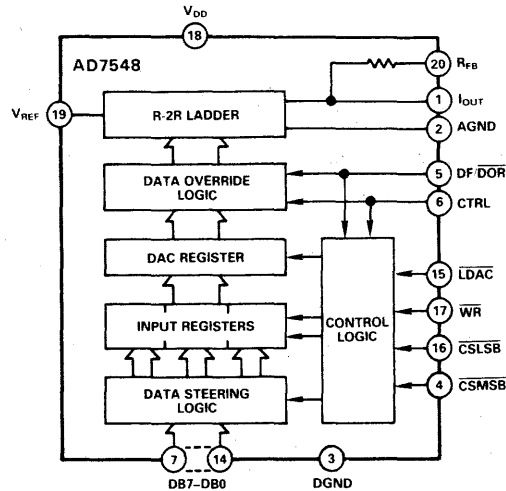
A separate load DAC control input allows the user the choice of updating the analog output coincident with loading new data to the DAC input register or at any time after the data loading event. This feature is especially important in multi-DAC systems where simultaneous update of all DACs is required.

The new Linear Compatible CMOS (LC<sup>2</sup> MOS) process used in the manufacture of the AD7548 allows precision thin-film linear circuitry and high-speed low-power CMOS logic to be integrated on the same small chip. The high-speed logic allows direct interfacing to most of the popular 8-bit microprocessors.

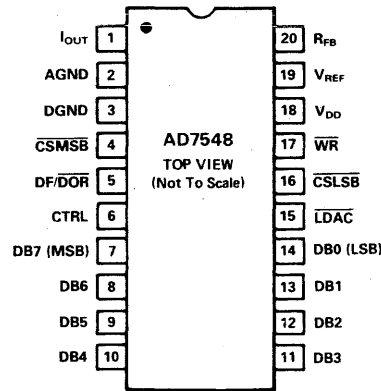
### ORDERING INFORMATION

Relative Accuracy $T_{MIN}$ to $T_{MAX}$	Full Scale Error $T_{MIN}$ to $T_{MAX}$	Temperature Range and Package		
		Plastic 0 to +70 $^{\circ}$ C	Cerdip -40 $^{\circ}$ C to +85 $^{\circ}$ C	Ceramic -55 $^{\circ}$ C to +125 $^{\circ}$ C
$\pm 11$ LSB	$\pm 6$ LSB	AD7548JN	AD7548AQ	AD7548SD
$\pm 12$ LSB	$\pm 3$ LSB	AD7548KN	AD7548BQ	AD7548TD

### AD7548 FUNCTIONAL BLOCK DIAGRAM



### PIN CONFIGURATION

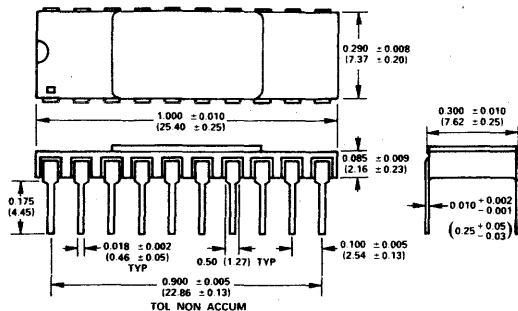


### MECHANICAL INFORMATION

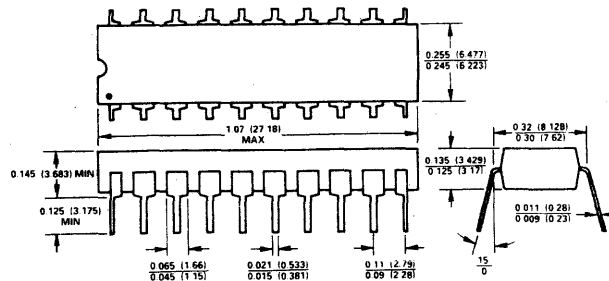
#### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

#### 20-PIN CERAMIC DIP (SUFFIX D)



#### 20-PIN PLASTIC DIP (SUFFIX N)



**FEATURES**

14-Bit Resolution  
200kHz Word Rates  
RZ Gated Output  
32-Pin DIP

**APPLICATIONS**

FDM/TDM Transmultiplexers  
Digital Signal Processing  
PCM Systems  
Digital Audio

**GENERAL DESCRIPTION**

The HDD-1409KM D/A converter is a voltage-output, 32-pin hybrid digital-to-analog converter complete with input registers, current-output D/A, switching circuits, and output amplifier. The unit is capable of converting 14-bit digital inputs into gated analog output voltages at update rates from dc through 200kHz.

Monolithic ICs and hybrid microelectronic packaging have been combined in a grounded metal case hybrid which provides cost, space, and power savings for the system designer. The HDD-1409 is a complete solution for converting high-resolution digital data into "clean" analog voltages and accomplishes it with maximum power dissipation of only 600 milliwatts.

The HDD-1409 D/A has been characterized with a companion A/D converter, the HAS-1409KM, to emphasize the superior ac performance which makes the A/D - D/A combination especially attractive for use in Frequency Division Multiplex/Time Division Multiplex (FDM/TDM) transmultiplexer systems. But the design concepts and versatility which are incorporated into it also make the HDD-1409 useful in Pulse Code Modulation (PCM) and other digital signal processing applications.

The analog output voltage range is  $\pm 5V$ ; output impedance is 600 ohms, ideal for filter matching. The D/A output operates in a return-to-zero (RZ) mode, which provides deglitching and allows selecting the optimum duty cycle for FDM/TDM and PCM applications.

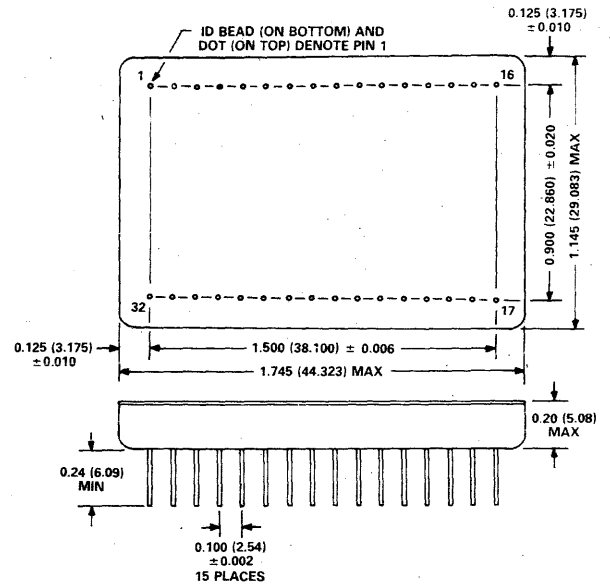
Small size, low power, and multiple functions in a single package make the HDD-1409 D/A converter attractive for a wide range of data processing uses.

**CAUTION:**

ESD (Electro-Static-Discharge) sensitive device. The digital control inputs are zener protected; however, permanent damage may occur on unconnected devices subject to high energy electrostatic fields. Unused devices must be stored in conductive foam or shunts. The foam should be discharged to the destination socket before devices are removed.

**OUTLINE DIMENSIONS**

Dimensions shown in inches and mm


**HDD-1409 PIN DESIGNATIONS**

PIN	FUNCTION	PIN	FUNCTION
1	SIGNAL	17	DIGITAL GROUND
2	SIGNAL	18	BIT 8
3	GATE	19	BIT 7
4	DIGITAL GROUND	20	BIT 6
5	ANALOG GROUND	21	BIT 5
6	N/C	22	BIT 4
7	+5V	23	BIT 3
8	DIGITAL GROUND	24	BIT 2
9	LOW STROBE	25	BIT 1 (MSB)
10	BIT 14 (LSB)	26	HIGH STROBE
11	BIT 13	27	-15V
12	BIT 12	28	+15V
13	BIT 11	29	ANALOG OUTPUT
14	BIT 10	30	ANALOG GROUND
15	BIT 9	31	ANALOG GROUND
16	DIGITAL GROUND	32	ANALOG GROUND

NOTES:  
WHEN USING WITH HAS-1409 A/D, CONNECT BIT 1 (MSB) OF A/D (PIN 22) TO BIT 1 (MSB) OF HDD-1409 D/A (PIN 25).

TO USE HDD-1409 D/A AS STAND-ALONE DEVICE WITH OFFSET BINARY INPUT, APPLY BIT 1 DIGITAL INPUT SIGNAL TO SIGNAL (PIN 1) AND CONNECT SIGNAL (PIN 2) TO BIT 1 (MSB) (PIN 25).

TO LOAD 14-BIT DATA FULLY PARALLEL, CONNECT HIGH STROBE (PIN 26) AND LOW STROBE (PIN 9) TOGETHER EXTERNALLY.



### FEATURES

- 5ns Settling Time
- 100MHz Update Rate
- 20mA Output Current
- ECL-Compatible
- 40MHz Multiplying Mode

### APPLICATIONS

- Raster Scan & Vector Graphic Displays
- High-Speed Waveform Generation
- Digital VCOs
- Ultra-Fast Digital Attenuators

### GENERAL DESCRIPTION

The Analog Devices AD9768SD D/A converter is a monolithic current-output converter which can accept 8 bits of ECL-level digital input voltages and convert them into analog signals at update rates as high as 100MHz. In addition to its use as a standard D/A converter, it can also be utilized as a two-quadrant multiplying D/A at multiplying bandwidths as high as 40MHz.

An inherently low glitch design is used, and the complementary current outputs are suitable for driving transmission lines directly. Nominal full-scale output is 20mA, which corresponds to a 1-volt drop across a 50Ω load, or ±1 volt across 100Ω returned to +1 volt. The actual output current is determined by the on-chip reference voltage ( $V_{REF} \approx -1.26V$ ) and an external current setting resistor,  $R_{SET}$ .

Full-scale output current  $I_{OUT}$  with digital "1" at all inputs is calculated with the equation:

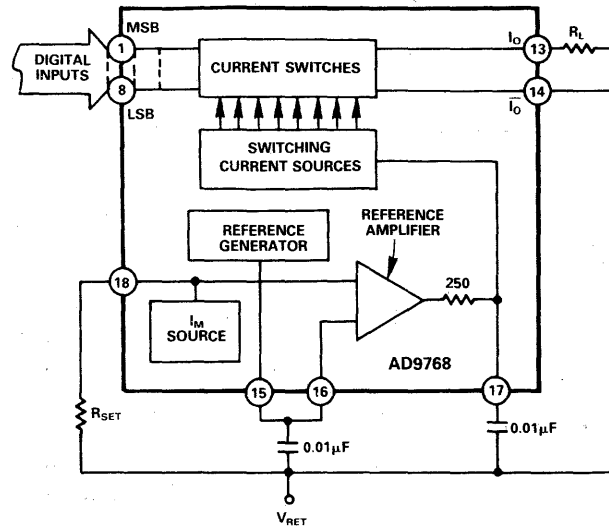
$$I_{OUT} = 4 \times \frac{V_{RET} - V_{REF}}{R_{SET}}$$

The setting resistor  $R_{SET}$  and the output load resistor should both have low temperature coefficients. A complementary  $I_{OUT}$  is also provided.

The reference voltage source is a modified bandgap type and is nominally -1.26 volts. This reference supply requires no external regulation. To reduce the possibility of noise generation and/or instability, pin 15 (REFERENCE OUT) can be decoupled using a high-quality ceramic chip capacitor. Stabilization of the internal loop amplifier is by a single capacitor connected from pin 17 (COMPENSATION) to ground. The minimum value for this capacitor is 3900pF, although a 0.01μF ceramic chip capacitor is recommended.

The incredible speed characteristics of the AD9768SD D/A converter make it attractive for a wide range of high-speed applications. The ability of the unit to operate as a two-quadrant multiplying D/A converter adds another dimension to its usefulness and makes the AD9768SD a truly versatile device.

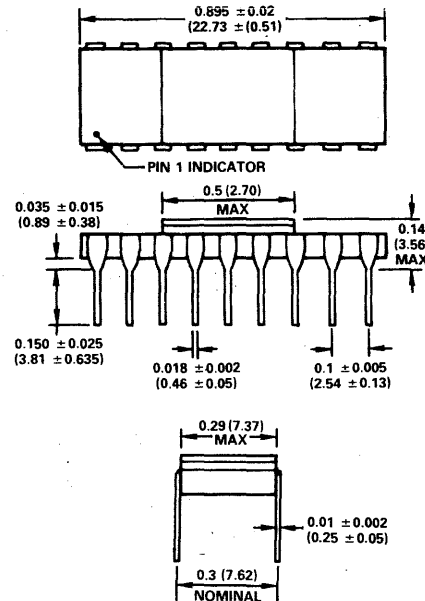
### AD9768 FUNCTIONAL BLOCK DIAGRAM



(Conventional Operation)

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



**FEATURE SELECTION CHART**

		GENERAL PURPOSE					FAST			μP BUS COMPATIBLE		SPECIAL PURPOSE		MULTIPLYING											
		AD1408	AD DAC0801	AD DAC080-V	AD DAC85-1	AD DAC85-V	AD DAC08	AD561	AD DAC100	AD558	AD7324	AD7323	AD7322	94B518MS DAC	AD7328	DUAL DAC	AD7323	AD7320	AD7320	AD7323	AD7321	AD7321	AD7341	AD7341A	AD7240
Resolution	8 Bits 10 Bits 12 Bits	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Accuracy	8 Bits 10 Bits 12 Bits	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Output Format	Current Voltage	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Internal Reference		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Multiplication Capability	2 Quadrant 4 Quadrant	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Logic	TTL Compatible with CMOS	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Input Coding	Binary BCD	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Second Source		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Single Power Supply		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Input Data Latch Structure	Serial Parallel	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Operating Temperature Ranges <sup>1</sup>	C = 0 to +70°C I = -25°C to +85°C M = -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Low Power		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Dice Availability		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•

NOTE:  
<sup>1</sup> C = Commercial I = Industrial M = Military

**GENERAL PURPOSE SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)**

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Setting Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD DAC-08	8 Bits	±0.19%	Monotonicity Guaranteed	50	135ns	M	I	TTL & CMOS	DIP
AD DAC-08A	8 Bits	±0.1%		50	135ns	M	I	CMOS	DIP
AD DAC-08C	8 Bits	±0.39%		80	135ns	C	I		DIP
AD DAC-08E	8 Bits	±0.19%		50	135ns	C	I		DIP
AD DAC-08H	8 Bits	±0.1%		50	135ns	C	I		DIP
AD1408-7	8 Bits	±0.39%	±2LSB	20 typ	250ns typ	C	I	TTL & CMOS	DIP
AD1408-8	8 Bits	±0.19%	±1LSB	20 typ	250ns typ	C	I		DIP
AD1408-9	8 Bits	±0.1%	±1/2LSB	20 typ	250ns typ	C	I		DIP
AD1508-8	8 Bits	±0.19%	±1LSB	20 typ	250ns typ	M	I		DIP
AD1508-9	8 Bits	±0.1%	±1/2LSB	20 typ	250ns typ	M	I		DIP
AD DAC80(Z)-CBI-V <sup>2</sup>	12 Bits	±0.012%	±3/4LSB	30	3μs typ	C	V	TTL	DIP
AD DAC80(Z)-CBI-I <sup>2</sup>	12 Bits	±0.012%	±3/4LSB	30	300ns typ	C	I		DIP
AD DAC80(Z)-CCD-V <sup>2</sup>	3 Digits	±0.024%	±1/2LSB	30	3μs typ	C	V		DIP
AD DAC80(Z)-CCD-I <sup>2</sup>	3 Digits	±0.024%	±1/2LSB	30	300ns typ	C	I		DIP
AD DAC85C-CBI-V	12 Bits	±0.012%	±1/2LSB	20	3μs typ	C	V	TTL	DIP
AD DAC85C-CBI-I	12 Bits	±0.012%	±1/2LSB	20	300ns typ	C	I		DIP
AD DAC85C-CCD-V	3 Digits	±0.024%	±1/2LSB	20	3μs typ	C	V		DIP
AD DAC85C-CCD-I	3 Digits	±0.024%	±1/2LSB	20	300ns typ	C	I		DIP
AD DAC85(LD)-CBI-V	12 Bits	±0.012%	±1/2LSB	20 (10)	3μs typ	I	V		DIP
AD DAC85(LD)-CBI-I	12 Bits	±0.012%	±1/2LSB	20 (10)	300ns typ	I	I		DIP
AD DAC85-CCD-V	3 Digits	±0.024%	±1/2LSB	20	3μs typ	I	V		DIP
AD DAC85-CCD-I	3 Digits	±0.025%	±1/2LSB	20	300ns typ	I	I		DIP
AD DAC85MIL-CBI-V	12 Bits	±0.012%	±1/2LSB	20	3μs typ	M	V		DIP
AD DAC85MIL-CBI-I	12 Bits	±0.012%	±1/2LSB	20	300ns typ	M	I		DIP
AD DAC100JD	10 Bits	±0.1%	GMOT <sup>3</sup>	±60	375ns	I	I	TTL & DTL	DIP
AD DAC100KD	10 Bits	±0.05%	GMOT <sup>3</sup>	±30	375ns	I	I		DIP
AD DAC100LD	10 Bits	±0.05%	GMOT <sup>3</sup>	±15	375ns	I	I		DIP
AD DAC100SD	10 Bits	±0.1%	GMOT <sup>3</sup>	±60	375ns	M	I		DIP
AD DAC100TD	10 Bits	±0.05%	GMOT <sup>3</sup>	±30	375ns	M	I		DIP
AD561JD	10 Bits	±0.05%	±1/2LSB	80	250ns typ	C	I	TTL & CMOS	DIP
AD561KD	10 Bits	±0.025%	±1/2LSB	30	250ns typ	C	I		DIP
AD561SD	10 Bits	±0.05%	±1/2LSB	60	250ns typ	M	I		DIP
AD561TD	10 Bits	±0.025%	±1/2LSB	30	250ns typ	M	I		DIP

NOTES

<sup>1</sup> C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C

<sup>2</sup> "Z" models specified for operation on ±12V supplies

<sup>3</sup> Guaranteed Monotonic Over Temperature

### MULTIPLYING SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD7523JN(AD)(SD)	8 Bits	±0.2%	Monotonicity	66/50/30	150ns typ	C/I/M	I	CMOS	DIP
AD7523KN(BD)	8 Bits	±0.1%	Guaranteed	66/50	150ns typ	C/I	I		DIP
AD7523LN(CD)	8 Bits	±0.05%		66/50	150ns typ	C/I	I		DIP
AD7520JN(JD)(SD)	10 Bits	±0.2%	±0.4%	10	500ns typ	C/I/M	I	CMOS &	DIP
AD7520KN(KD)(TD)	10 Bits	±0.1%	±0.2%	10	500ns typ	C/I/M	I	TTL	DIP
AD7520LN(LD)(UD)	10 Bits	±0.05%	±0.1%	10	500ns typ	C/I/M	I		DIP
AD7530JN(JD)	10 Bits	±0.2%	±0.4%	10	500ns typ	C/I	I	CMOS &	DIP
AD7530KN(KD)	10 Bits	±0.1%	±0.2%	10	500ns typ	C/I	I	TTL	DIP
AD7530LN(LD)	10 Bits	±0.05%	±0.1%	10	500ns typ	C/I	I		DIP
AD7533JN(AD)(SD)	10 Bits	±0.2%	±0.4%	22/16/10	600ns	C/I/M/M	I	CMOS &	DIP
AD7533KN(BD)(TD)	10 Bits	±0.1%	±0.2%	22/16/10	600ns	C/I/M/M	I	CMOS &	DIP
AD7533LN(LD)(UD)	10 Bits	±0.05%	±0.1%	22/16/10	600ns	C/I/M/M	I	CMOS &	DIP
AD7521JN(JD)(SD)	12 Bits	±0.2%	±0.4%	10	500ns typ	C/I/M	I	TTL	DIP
AD7521KN(KD)(TD)	12 Bits	±0.1%	±0.2%	10	500ns typ	C/I/M	I	CMOS &	DIP
AD7521LN(LD)(UD)	12 Bits	±0.05%	±0.1%	10	500ns typ	C/I/M	I	TTL	DIP
AD7531JN(JD)	12 Bits	±0.2%	±0.4%	10	500ns typ	C/I	I		DIP
AD7531KN(KD)	12 Bits	±0.1%	±0.2%	10	500ns typ	C/I	I	CMOS &	DIP
AD7531LN(LD)	12 Bits	±0.05%	±0.1%	10	500ns typ	C/I	I	TTL	DIP

NOTE  
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C

### µP COMPATIBLE SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD558JD(SD)	8 Bits	±1/2LSB	Monotonicity	±2LSB	800ns typ	C/M	V	TTL &	DIP
AD558KD(TD)	8 Bits	±1/4LSB	Guaranteed	±2LSB	800ns typ	C/M	V	CMOS	DIP
AD7524JN(AD)(SD)	8 Bits	±0.2%	Monotonicity	10	100ns	C/I/M	I	TTL &	DIP
AD7524KN(BD)(TD)	8 Bits	±0.1%	Guaranteed	10	100ns	C/I/M	I	CMOS	DIP
AD7524LN(CD)(UD)	8 Bits	±0.05%		10	100ns	C/I/M	I		DIP
AD7522JN(JD)(SD)	10 Bits	±0.2%	±0.4%	10	500ns typ	C/I/M	I	TTL &	DIP
AD7522KN(KD)(TD)	10 Bits	±0.1%	±0.2%	10	500ns typ	C/I/M	I	CMOS	DIP
AD7522LN(LD)(UD)	10 Bits	±0.05%	±0.1%	10	500ns typ	C/I/M	I		DIP
<b>DUAL MULTIPLYING</b>									
AD7528JN(AQ)(SD)	8 Bits	±1LSB	±1LSB	±35 <sup>2</sup>	200ns <sup>2</sup>	C/I/M	I	TTL &	DIP
AD7528KN(LN)	8 Bits	±1/2LSB	±1LSB	±35 <sup>2</sup>	200ns <sup>2</sup>	C	I	CMOS	DIP
AD7528BQ(CQ)	8 Bits	±1/2LSB	±1LSB	±35 <sup>2</sup>	200ns <sup>2</sup>	I	I	TTL &	DIP
AD7528TD(UD)	8 Bits	±1/2LSB	±1LSB	±35 <sup>2</sup>	200ns <sup>2</sup>	M	I	CMOS	DIP
<b>SYSTEMS DAC<sup>3</sup></b>									
AD7527KN(BD)(TD)	10 Bits	±1LSB	±2LSB	5	—	C/I/M	I	CMOS	DIP
AD7527LN(CD)(UD)	10 Bits	±0.5LSB	±1LSB	5	—	C/I/M	I	CMOS	DIP
AD7527GLN(GCD)(GUD)	10 Bits	±0.5LSB	±1LSB	5	—	C/I/M	I	CMOS	DIP

NOTES  
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C  
<sup>2</sup>+15V supply  
<sup>3</sup>Has double-buffered latches, data readback, data override, L/R - justified data, increment/decrement

### HIGH PERFORMANCE SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD563JD-BIN/BCD	12 Bits/3 Digits	±1/2LSB	±1/2LSB	50	1.5µs typ	C	I	TTL &	DIP
AD563KD(SD)(TD)-BIN/BCD	12 Bits/3 Digits	±1/4LSB	±1/2LSB	20/30/10	1.5µs typ	C/M	I	CMOS	DIP
AD565JD(SD)	12 Bits	±1/2LSB	±3/4LSB	30	400ns	C/M/M	I	TTL	DIP
AD565KD(TD)	12 Bits	±1/4LSB	±1/2LSB	20/15/15	400ns	C/M/M	I		DIP
AD565AJD(SD)	12 Bits	±1/2LSB	±3/4LSB	50/30/30	250ns	C/M/M	I	TTL	DIP
AD565AKD(TD)	12 Bits	±1/4LSB	±1/2LSB	20/15/15	250ns	C/M/M	I	TTL	DIP
AD DAC87-CBI-I	12 Bits	±1/2LSB	±3/4LSB	25	3µs	M/M	I	TTL &	DIP
AD DAC87-CBI-V	12 Bits	±1/2LSB	±3/4LSB	25	300ns	M/M	V	CMOS	DIP
AD370JN(JD)	12 Bits	±1LSB			35µs	C/C	V	TTL &	DIP
AD370KN(KD)(SD)	12 Bits	±1/2LSB	Monotonic		35µs	C/C/M/M	V	CMOS	DIP
AD371JN(JD)	12 Bits	±1LSB			35µs	C/C	V	TTL &	DIP
AD371KN(KD)(SD)	12 Bits	±1/2LSB	Monotonic		35µs	C/C/M/M	V	CMOS	DIP

NOTES  
<sup>1</sup>C = 0 to +70°C, M = -55°C to +125°C

## HIGH PERFORMANCE DACS FEATURE SELECTION CHART

		HIGH PERFORMANCE										MULTIPLYING		
							μP COMPATIBLE			FAST				
		AD1463	AD DAC87	AD1770	AD1771	AD DAC71 DAC72	AD17346	AD1860	AD1507	AD1607	AD1865	AD1865A	AD1866	AD1866A
Resolution	0.375dB Steps to 88.5dB													
	1.5dB Steps to 88.5dB													
	8 Bits													
	10 Bits													
	12 Bits	•	•	•	•	•	•	•	•	•	•	•	•	•
Output Format	Current	•	•	•	•	•	•	•	•	•	•	•	•	•
	Voltage		•	•	•	•	•	•	•	•	•	•	•	•
Internal Reference Multiplication Capability	2 Quadrant	•	•	•	•	•	•	•	•	•	•	•	•	•
	4 Quadrant						•						•	
Logic Compatible with	TTL	•	•	•	•	•	•	•	•	•	•	•	•	•
	CMOS	•	•	•	•	•	•	•	•	•	•	•	•	•
Input Code	Binary	•	•	•	•	•	•	•	•	•	•	•	•	•
	BCD	•	•	•	•	•	•	•	•	•	•	•	•	•
Second Source		•	•	•	•	•	•	•	•	•	•	•	•	
Single Power Supply														
Input Data Latch Structure	Serial													
	Parallel						•	•	•	•	•	•	•	•
Operating Temperature Ranges <sup>1</sup>	C = 0 to +70°C	•	•	•	•	•	•	•	•	•	•	•	•	•
	I = -25°C to +85°C	•	•	•	•	•	•	•	•	•	•	•	•	•
	M = -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•	•
Low Power				•	•		•							
Input/Output Isolation														
Dice Availability								•			•		•	

NOTE

<sup>1</sup>C = Commercial I = Industrial M = Military

## HIGH PERFORMANCE DACS FEATURE SELECTION CHART (Continued)

		HIGH PERFORMANCE										SPECIAL PURPOSE			
		MULTIPLYING					μP BUS COMPATIBLE			LOG DAC™ ATTENUATORS		QUAD DAC			
		AD1862	AD1741	AD1741A	AD1733	AD1742	AD1743	AD1744	AD1745	AD1748	AD1715	AD1711	AD1710	AD1718	AD1700
Resolution	0.1dB Steps to 19.9dB									•	•	•	•		
	0.375dB Steps to 88.5dB														
	1.5dB Steps to 88.5dB				•										
	8 Bits														
Output Format	Current	•	•	•	•	•	•	•	•	•	•	•	•	•	
	Voltage		•	•	•	•	•	•	•	•	•	•	•	•	
Internal Reference Multiplication Capability	2 Quadrant	•	•	•	•	•	•	•	•	•	•	•	•	•	
	4 Quadrant													•	
Logic Compatible with	TTL	•	•	•	•	•	•	•	•	•	•	•	•	•	
	CMOS	•	•	•	•	•	•	•	•	•	•	•	•	•	
Input Code	Binary	•	•	•	•	•	•	•	•	•	•	•	•	•	
	BCD	•	•	•	•	•	•	•	•	•	•	•	•	•	
Second Source		•	•	•											
Single Power Supply										•	•		•		
Input Data Latch Structure	Serial														
	Parallel					•	•	•	•	•	•	•	•	•	
Operating Temperature Ranges <sup>2</sup>	C = 0 to +70°C	•	•	•	•	•	•	•	•	•	•	•	•	•	
	I = -25°C to +85°C	•	•	•	•	•	•	•	•	•	•	•	•	•	
	M = -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•	•	
Low Power															
Input/Output Isolation															
Dice Availability			•			•	•								

NOTE

<sup>2</sup>Six-Stage FIFO Input Register <sup>3</sup>C = Commercial I = Industrial M = Military



**μP BUS COMPATIBLE**
**SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)**

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD567J(S)	12 Bits	±1/2LSB	±3/4LSB	50/30/30	500ns	C/M/M	I	TTL & CMOS	DIP
AD567K(T)	12 Bits	±1/4LSB	±1/2LSB	20	500ns	C/M/M	I	CMOS	DIP
AD667J(S)	12 Bits	±1/2LSB	±3/4LSB	50/30	2μs <sup>2</sup>	C/M	V	TTL	DIP
AD667K	12 Bits	±1/2LSB	±1/2LSB	40	2μs <sup>2</sup>	C	V	TTL	DIP
AD3860K	12 Bits	±1/2LSB	±1LSB	±10	5μs <sup>2</sup>	C	V	TTL	DIP
AD3860S	12 Bits	±1/2LSB	GMOT <sup>3</sup>	±10	5μs <sup>2</sup>	M	V	TTL	DIP
AD7542JN(AD)(SD)	12 Bits	±1LSB	±0.048%	5	2μs	C/I/I/ M/M	I	CMOS	DIP
AD7542KN(BD)(TD)	12 Bits	±1/2LSB	±0.024%	5	2μs	C/I/I/ M/M	I	CMOS	DIP
AD7543JN(AD)(SD)	12 Bits	±1LSB	±0.048%	5	2μs	C/I/I/ M/M	I	CMOS	DIP
AD7543KN(BD)(TD)	12 Bits	±1/2LSB	±0.024%	5	2μs	C/I/I/ M/M	I	CMOS	DIP
AD7544JN(AD)(SD) <sup>4</sup>	12 Bits	±1LSB	±0.048%	5	2μs	C/I/M	I	CMOS	DIP
AD7544KN(BD)(TD) <sup>4</sup>	12 Bits	±1/2LSB	±0.024%	5	2μs	C/I/M	I	CMOS	DIP
AD7544GKN(GBD)(GTD) <sup>4</sup>	12 Bits	±1/2LSB	±0.024%	5	2μs	C/I/M	I	CMOS	DIP
AD7545J(A)(S)	12 Bits	±2LSB	±4LSB	±5 <sup>5</sup>		C/I/M	I	TTL & CMOS	DIP
AD7545K(B)(T)	12 Bits	±1LSB	±1LSB	±5 <sup>5</sup>		C/I/M	I	CMOS	DIP
AD7545L(C)(U)	12 Bits	±1/2LSB	±1LSB	±5 <sup>5</sup>		C/I/M	I		DIP
AD7545GL(GC)(GU) <sup>6</sup>	12 Bits	±1/2LSB	±1LSB	±5 <sup>5</sup>		C/I/M	I		DIP
AD7548J(A)(S)	12 Bits	±1LSB	±1LSB	±5 <sup>5</sup>		C/I/M	I	TTL	DIP
AD7548K(B)(T)	12 Bits	±1/2LSB	±1/2LSB	±5 <sup>5</sup>		C/I/M	I	TTL	DIP
AD7546JN(AD)	16 Bits	±0.05% FSR	±0.006% FSR	±2	10μs typ <sup>7</sup>	C/I	V	TTL & CMOS	DIP
AD7546KN(BD)	16 Bits	±0.012% FSR	GMOT <sup>3</sup>	±2	10μs typ <sup>7</sup>	C/I	V	CMOS	DIP
<b>LOG DAC™ STEP ATTENUATORS</b>									
AD7111K(B)(T)	0.375dB	30dB ± 0.17dB	GMOT <sup>8</sup>			C/I/M	I	CMOS	DIP
AD7111L(C)(U)	0.375dB	36dB ± 0.17dB	GMOT <sup>9</sup>			C/I/M	I	CMOS	DIP
AD7115K(B)(T)	0.1dB	±0.05dB	Monotonic			C/I/M	I	TTL & CMOS	DIP
<b>QUAD DAC (FOUR DACS IN ONE PACKAGE)</b>									
AD390J(D)(SD)	12 Bits	±3/4LSB	GMOT <sup>3</sup>	±40/±10 <sup>10</sup>	8μs	C/M	V	TTL & CMOS	DIP
AD390K(D)(TD)	12 Bits	±1/2LSB	GMOT <sup>3</sup>	±10/±5 <sup>10</sup>	8μs	C/M	V	CMOS	DIP

**NOTES**
<sup>1</sup> C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C

<sup>2</sup> To 0.01% for 10-V Step

<sup>3</sup> Guaranteed Monotonic Over Temperature

<sup>4</sup> Six-Stage FIFO Input Register

<sup>5</sup> 5 Volt Supply

<sup>6</sup> 2LSB max gain error (multiplying DAC)

<sup>7</sup> To 0.00076% of final value

<sup>8</sup> 0 to -48dB at 0.375dB, 0 to -60dB at 0.75dB, 0 to -72dB at 1.5dB

<sup>9</sup> 0 to -54dB at 0.375dB, 0 to -72dB at 0.75dB, 0 to -88.5dB at 1.5dB

<sup>10</sup> Internal/External reference

**HIGH RESOLUTION**
**SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)**

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD7546JN(AD)	16 Bits	±0.05%	±0.006%	±2	10μs typ <sup>2</sup>	C/I	V	CMOS or	DIP
AD7546KN(BD)	16 Bits	±0.012%	±0.0015%	±2	10μs typ <sup>2</sup>	C/I	V	TTL	DIP
AD DAC71	16 Bits	±0.003%	Note 3	±15	10/1μs <sup>4</sup>	C	V/I	TTL	DIP <sup>5</sup>
AD DAC71H	16 Bits	±0.003%	Note 3	±15	10/1μs <sup>4</sup>	C	V/I	TTL	DIP <sup>5</sup>
AD DAC72C	16 Bits	±0.003%	Note 3	±15	10/1μs <sup>4</sup>	C	V/I	TTL	DIP <sup>5</sup>
AD DAC72	16 Bits	±0.003%	Note 6	±15 <sup>6</sup>	10/1μs <sup>4</sup>	I	V/I	TTL	DIP <sup>5</sup>

**NOTES**
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C

<sup>2</sup>to 0.00076% of final value

<sup>3</sup>Guaranteed 14-bit monotonic 0 to +50°C

<sup>4</sup>Voltage/current, to ±0.003% FSR

<sup>5</sup>AD DAC71 polymer sealed, AD DAC71H and AD DAC72 hermetic

<sup>6</sup>Guaranteed 14-bit monotonic 0 to +70°C, Gain tempco ±15ppm/°C, T<sub>min</sub> to +25°C, ±7ppm/°C, +25°C to T<sub>max</sub>
**SPECIAL PURPOSE**
**SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)**

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
<b>LOG DAC™ STEP ATTENUATORS</b>									
AD7110KN	1.5dB	±0.7dB, 0 to -48dB	Monotonic	NA	NA	0 to +50°C	I	CMOS	DIP
AD7118KN(BD)(TD)	1.5dB	±0.75dB, 0 to -42dB	Monotonic	NA	NA	C/I/M	I	CMOS & TTL	DIP
AD7118LN(CD)(UD)	1.5dB	±0.7dB, 0 to -48dB	Monotonic	NA	NA	C/I/M	I	CMOS & TTL	DIP

These devices are digitally controlled attenuators for use in high performance audio systems and wide dynamic range applications.

**NOTE**
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C

**MULTIPLYING**

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Settling Time	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD562KD(AD)-BIN	12 Bits	±1/2LSB	±1/2LSB	3	1.5μs typ	C/I	I	TTL &	DIP
AD562SD-BIN	12 Bits	±1/4LSB	±1/2LSB	3	1.5μs typ	M	I	CMOS	DIP
AD562KD(AD)-BCD	3 Digits	±1/2LSB	±1/2LSB	3	1.5μs typ	C/I	I		DIP
AD562SD-BCD	3 Digits	±1/10LSB	±1/2LSB	3	1.5μs typ	M	I		DIP
AD566JD(SD)	12 Bits	±1/2LSB	±3/4LSB	10	400ns	C/M/M	I	TTL	DIP
AD566KD(TD)	12 Bits	±1/4LSB	±1/2LSB	3	400ns	C/M/M	I		DIP
AD566AJD(SD)	12 Bits	±1/2LSB	±3/4LSB	10	350ns	C/M/M	I	TTL	DIP
AD566AKD(TD)	12 Bits	±1/4LSB	±1/2LSB	3	350ns	C/M/M	I	TTL	DIP
AD7541JN(AD)	12 Bits	±1LSB	±2LSB	10	1μs	C/I/M/M	I	TTL &	DIP
AD7541KN(BD)	12 Bits	±1/2LSB	±1LSB	10	1μs	C/I/M/M	I	CMOS	DIP
AD7541AJN(AQ)	12 Bits	±1LSB	±1LSB	5	0.6μs typ <sup>2</sup>	C/I	I	TTL &	DIP
AD7541AKN(BQ)	12 Bits	±1/2LSB	±1/2LSB	5	0.6μs typ <sup>2</sup>	C/I	I	CMOS	DIP
AD7541ASD	12 Bits	±1LSB	±1LSB	5	0.6μs typ <sup>2</sup>	M/M	I		DIP
AD7541ATD	12 Bits	±1/2LSB	±1/2LSB	5	0.6μs typ <sup>2</sup>	M/M	I		DIP
AD7240J(A)(S)	12 Bits	+1/2, -1 1/4	±1/2LSB	6	900ns	C/I/M	V	TTL	DIP
AD7240K(B)(T)	12 Bits	+1/2, -1LSB	±1/2LSB	6	900ns	C/I/M	V	TTL	DIP
AD7525KN(BD)(TD)	3 1/2 Digits	±1LSB	±1LSB	25	1μs	C/I/M/M	I	CMOS	DIP
AD7525LN(CD)(UD)	3 1/2 Digits	±1/2 LSB	±1/2LSB	25	1μs	C/I/M/M	I		DIP

<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C

## ULTRA FAST/VIDEO DACS FEATURE SELECTION CHART

		CURRENT OUT										
		HDC-0405	HDC-0605	HDC-0805	HDS-0820	HDS-0810E	HDD-0810	HDD-0810C	HDS-1025	HDS-1015E	HDD-1015	AD9768
Settling Time to Resolution in Bits	<550ns											
	<210ns											
	<75ns											
	<21ns											
	<8ns	•	•	•	•	•	•	•	•	•	•	•
Resolution	4 Bits	•										
	6 Bits		•									
	8 Bits			•	•	•	•	•				•
	10 Bits								•	•	•	
	11 Bits											
	12 Bits											
Application	General Purpose	•	•	•	•	•	•	•	•	•	•	•
	Lowest Glitch	•	•	•				•				
	Multiplying Composite Video	•	•	•				•				•
Output Format	Current	•	•	•	•	•	•	•	•	•	•	•
	Voltage											
Binary Logic	TTL				•				•			
	ECL	•	•	•		•	•	•		•	•	•
	Latched	•	•	•								
Single Power Supply		•	•	•		•	•	•		•	•	
Operating Temperature Ranges	0 to +70°C	•	•	•	•	•	•	•	•	•	•	•
	-25°C to +85°C				•	•	•	•	•	•	•	•
	-55°C to +125°C					•	•	•	•	•	•	•

## ULTRA FAST/VIDEO DACS FEATURE SELECTION CHART (Continued)

		CURRENT OUT				VOLTAGE OUT		
		HDD-1015C	HDS-1250	HDS-1240E	HDI-0802	HDI-1003	HDI-1205	HDD-1206
Settling Time to Resolution in Bits	<2µs						•	•
	<210ns							
	<75ns			•	•			
	<21ns	•						
	<8ns							
Resolution	4 Bits							
	6 Bits							
	8 Bits				•			
	10 Bits	•				•		
	11 Bits							
	12 Bits			•	•		•	•
Application	General Purpose		•	•	•	•	•	•
	Lowest Glitch							•
	Multiplying Composite Video	•						
Output Format	Current	•	•	•				
	Voltage				•	•	•	•
Binary Logic	TTL		•		•	•	•	•
	ECL	•		•				
	Latched	•						•
Single Power Supply		•						
Operating Temperature Ranges	0 to +70°C	•	•	•	•	•	•	•
	-25°C to +85°C	•						
	-55°C to +125°C			• <sup>1</sup>	•	•	•	•

NOTE  
<sup>1</sup>-55°C to +100°C

**ULTRA FAST/VIDEO**
**SPECIFICATIONS (maximum @ +25°C unless otherwise noted as typical)**

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C Typical	Settling Time ns Typical	Temp Range <sup>1</sup>	I/V Out	Input Logic	Package
AD9768	8 Bits	±1/2LSB	GMOT <sup>2</sup>	80	5	M	I	ECL	DIP
HDG-0405 <sup>3,4</sup>	4 Bits	±1/2LSB	GMOT <sup>2</sup>	25	4	C	I	ECL	DIP
HDG-0605 <sup>3,4</sup>	6 Bits	±1/2LSB	GMOT <sup>2</sup>	25	6	C	I	ECL	DIP
HDG-0805 <sup>3,4</sup>	8 Bits	±1/2LSB	GMOT <sup>2</sup>	25	8	C	I	ECL	DIP
HDS-0820(0820M)	8 Bits	±1/4LSB	GMOT <sup>2</sup>	30	20	C/M	I	TTL	DIP
HDS-0810E(0810EM)	8 Bits	±1/4LSB	GMOT <sup>2</sup>	80	10	C/M	I	ECL	DIP
HDD-0810(0810M) <sup>4</sup>	8 Bits	±1/4LSB	GMOT <sup>2</sup>	80	10	C/M	I	ECL	DIP
HDD-0810C(0810CM) <sup>3,4</sup>	8 Bits	±1/4LSB	GMOT <sup>2</sup>	80	10	C/M	I	ECL	DIP
HDS-1025(1025M)	10 Bits	±1/2LSB	GMOT <sup>2</sup>	30	25	C/M	I	TTL	DIP
HDS-1015E(1015EM)	10 Bits	±1/2LSB	GMOT <sup>2</sup>	80	15	C/M	I	ECL	DIP
HDD-1015(1015M) <sup>4</sup>	10 Bits	±1/2LSB	GMOT <sup>2</sup>	80	15	C/M	I	ECL	DIP
HDD-1015C(1015CM) <sup>3,4</sup>	10 Bits	±1/2LSB	GMOT <sup>2</sup>	80	15	C/M	I	ECL	DIP
HDS-1250(1250M)	12 Bits	±1/2LSB	GMOT <sup>2</sup>	30	35	C/M	I	TTL	DIP
HDS-1240E(1240EM)	12 Bits	±1/2LSB	GMOT <sup>2</sup>	25	35	C/MR	I	ECL	DIP
HDD-1206JW	12 Bits	±1/2LSB	GMOT <sup>2</sup>	40	60 <sup>5</sup>	C	V	TTL	DIP
HDD-1206SM	12 Bits	±1/2LSB	GMOT <sup>2</sup>	40	60 <sup>5</sup>	M	V	TTL	DIP
HDH-0802(0802M)	8 Bits	±1/4LSB	GMOT <sup>2</sup>	30	200	C/M	V	TTL	DIP
HDH-1003(1003M)	10 Bits	±1/2LSB	GMOT <sup>2</sup>	30	300	C/M	V	TTL	DIP
HDH-1205(1205M)	12 Bits	±1/2LSB	GMOT <sup>2</sup>	30	500	C/M	V	TTL	DIP

**NOTES**
<sup>1</sup> C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C, MR = -55°C to +100°C

<sup>2</sup> Guaranteed Monotonic Over Temperature

<sup>3</sup> Composite video

<sup>4</sup> Latched input

<sup>5</sup> 1LSB settling

# Fast, Complete 10-Bit A/D Converter with Microprocessor Interface

## AD573\*

### FEATURES

Complete 10-Bit A/D Converter with Reference, Clock and Comparator  
Full 8- or 16-Bit Microprocessor Bus Interface  
Fast Successive Approximation Conversion – 20 $\mu$ s  
No Missing Codes Over Temperature  
Operates on +5V and –12V to –15V Supplies  
Low Cost Monolithic Construction

### PRODUCT DESCRIPTION

The AD573 is a complete 10-bit successive approximation analog to digital converter consisting of a DAC, voltage reference, clock, comparator, successive approximation register (SAR) and 3 state output buffers—all fabricated on a single chip. No external components are required to perform a full accuracy 10-bit conversion in 20 $\mu$ s.

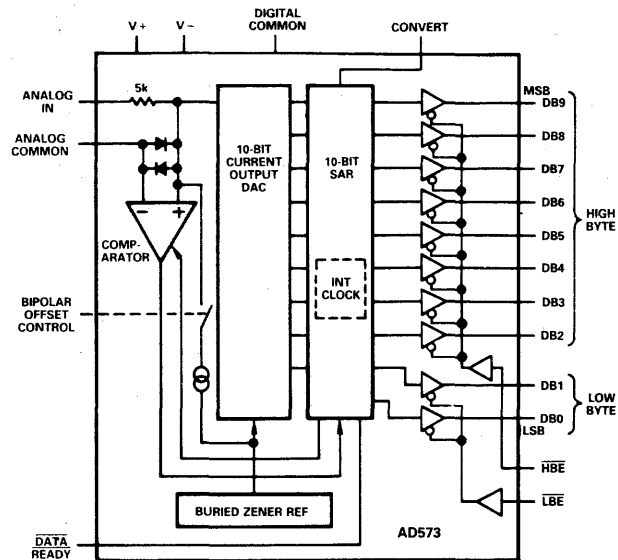
The AD573 incorporates the most advanced integrated circuit design and processing technology available today. The successive approximation function is implemented with I<sup>2</sup>L (integrated injection logic). Laser trimming of the high stability SiCr thin film resistor ladder network at the wafer stage (LWT) insures high accuracy, which is maintained with a temperature compensated sub-surface Zener reference.

Operating on supplies of +5V and –12V to –15V, the AD573 will accept analog inputs of 0 to +10V or –5V to +5V. A positive pulse on the CONVERT line initiates the 20 $\mu$ s conversion cycle. DATA READY indicates completion of the conversion. HIGH BYTE ENABLE (HBE) and LOW BYTE ENABLE (LBE) control the 8-bit and 2-bit three state output buffers.

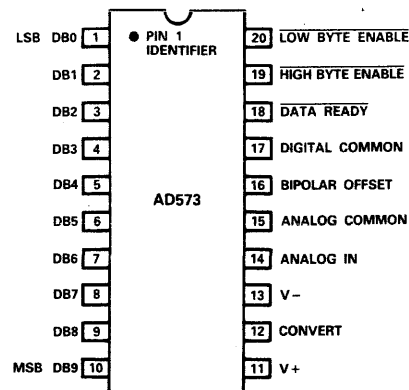
The AD573 is available in two versions for the 0 to +70°C temperature range, the AD573J and AD573K. The AD573S guarantees 10-bit relative accuracy and no missing codes from –55°C to +125°C.

Two package configurations are offered. The AD573J and AD573K are also available in a 20-pin plastic DIP. All versions are offered in a 20-pin hermetically sealed ceramic DIP.

### AD573 FUNCTIONAL BLOCK DIAGRAM



### AD573 PINOUT



### AD573 ORDERING GUIDE

Model	Package Option	Temperature Range	Relative Accuracy
AD573JN	20-Pin Plastic DIP	0 to +70°C	± 1LSB max
AD573KN	20-Pin Plastic DIP	0 to +70°C	± 1/2LSB max
AD573JD	20-Pin Ceramic DIP	0 to +70°C	± 1LSB max
AD573KD	20-Pin Ceramic DIP	0 to +70°C	± 1/2LSB max
AD573SD	20-Pin Ceramic DIP	–55°C to +125°C	± 1LSB max

\*Protected by U.S. Patent Nos. 3,940,760; 4,213,806 and 4,136,349



# Fast, Complete 12-Bit A/D Converter with Microprocessor Interface

## AD574A

### FEATURES

- Complete 12-Bit A/D Converter with Reference and Clock
- Full 8- or 16-Bit Microprocessor Bus Interface
- 250ns Bus Access Time
- Guaranteed Linearity Over Temperature
  - 0 to +70°C – AD574AJ, AK, AL
  - 55°C to +125°C – AD574AS, AT, AU
- No Missing Codes Over Temperature
- Fast Successive Approximation Conversion – 25µs
- Buried Zener Reference for Long-Term Stability and Low Gain T.C. 10ppm/°C max AD574AL  
12.5ppm/°C max AD574AU
- Low Profile 28-Pin Ceramic DIP
- Low Power: 390mW
- Low Cost

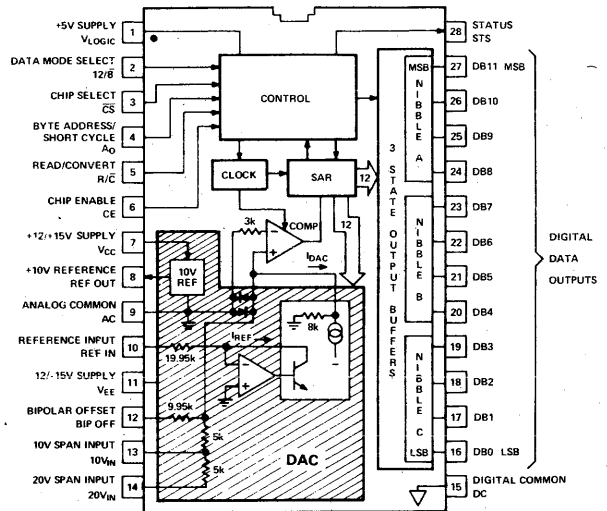
### PRODUCT DESCRIPTION

The AD574A is a complete 12-bit successive-approximation analog-to-digital converter with 3-state output buffer circuitry for direct interface to an 8-, 12- or 16-bit microprocessor bus. The AD574A design is implemented with two LSI chips each containing both analog and digital circuitry, resulting in the maximum performance and flexibility at the lowest cost.

One chip is the high performance AD565A 12-bit DAC and voltage reference. It contains the high speed current output switching circuitry, laser-trimmed thin film resistor network, low T.C. buried zener reference and the precision input scaling and bipolar offset resistors. This chip is laser-trimmed at the wafer stage (LWT) to adjust ladder network linearity, voltage reference tolerance and temperature coefficient, and the calibration accuracy of input scaling and bipolar offset resistors.

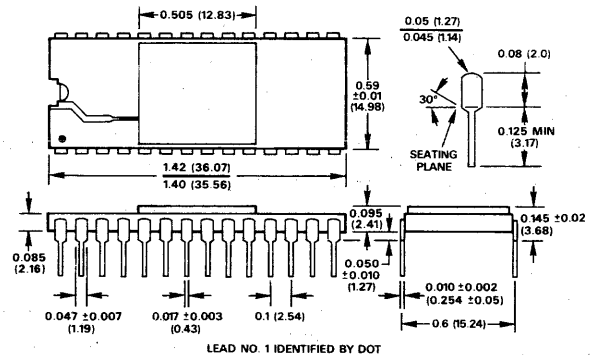
The second chip uses the proven LCI (linear-compatible integrated injection logic) process to provide the low-power I<sup>2</sup>L successive-approximation register, converter control circuitry, clock, bus interface, and the high performance latching comparator. The precision, low-drift comparator is adjusted for initial input offset error at the wafer stage by the "zener-zap" technique which trims the comparator input stage to 1/10 LSB typical error. This form of trimming, while cumbersome for complex ladder networks, is an attractive alternative to thin film resistor trimming for a simple offset adjustment and eliminates the need for thin film processing for this portion of the circuitry.

### AD574A FUNCTIONAL BLOCK DIAGRAM AND PINOUT



### AD574A PACKAGE OUTLINE

Dimensions shown in inches and (mm).



The AD574A is available in six different grades. The AD574AJ, AK, and AL grades are specified for operation over the 0 to +70°C temperature range. The AD574AS, AT, and AU are specified for the -55°C to +125°C range. All grades are packaged in a low-profile, 0.600 inch wide, 28-pin hermetically-sealed ceramic DIP.

### AD574A ORDERING GUIDE

Model	Temp. Range	Linearity Error Max (T <sub>min</sub> to T <sub>max</sub> )	Resolution No Missing Codes (T <sub>min</sub> to T <sub>max</sub> )	Max Full Scale T.C. (ppm/°C)
AD574AJD	0 to +70°C	±1LSB	11 Bits	50.0
AD574AKD	0 to +70°C	±1/2LSB	12 Bits	27.0
AD574ALD	0 to +70°C	±1/2LSB	12 Bits	10.0
AD574ASD	-55°C to +125°C	±1LSB	11 Bits	50.0
AD574ATD	-55°C to +125°C	±1LSB	12 Bits	25.0
AD574AUD	-55°C to +125°C	±1LSB	12 Bits	12.5

### FEATURES

- 10-Bit Plus Sign Resolution
- No Missed Codes Over Full Temperature Range
- Conversion Time 80µs
- Differential Analog Voltage Inputs, ±10V Range
- Serial and Parallel Data Outputs
- Easy Interface to Most Microprocessors
- Internal Clock Oscillator
- Single Supply Operation for Positive-Only Signals
- Monolithic Construction

### GENERAL DESCRIPTION

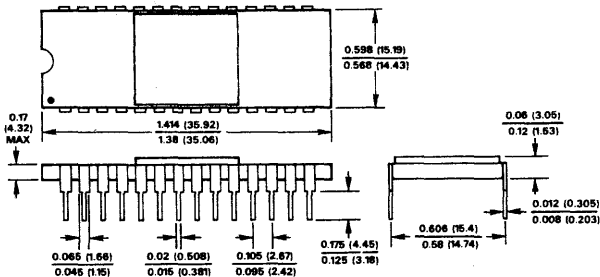
The AD7571 is a high speed, low cost 10-bit plus sign CMOS A/D converter which uses the successive approximation technique to provide a conversion time of 80µs. The device is designed for easy microprocessor interface allowing full parallel or double byte reading over three-state outputs. Conversion results are also available in serial form allowing opto-isolated operation using as few as two wires for the interconnect.

A new differential analog input configuration is used in the AD7571, increasing the common-mode rejection performance and allowing the analog zero input voltage to be offset from true analog ground. Analog input voltage range is ±10V using a single positive reference. With positive-only input signals the AD7571 can be operated from a single positive power supply.

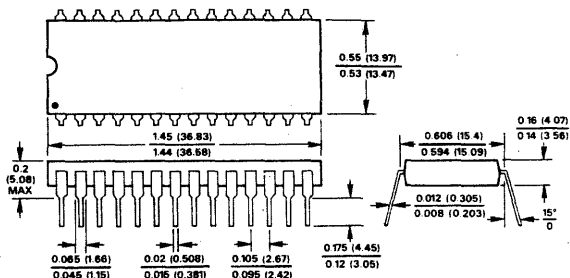
### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

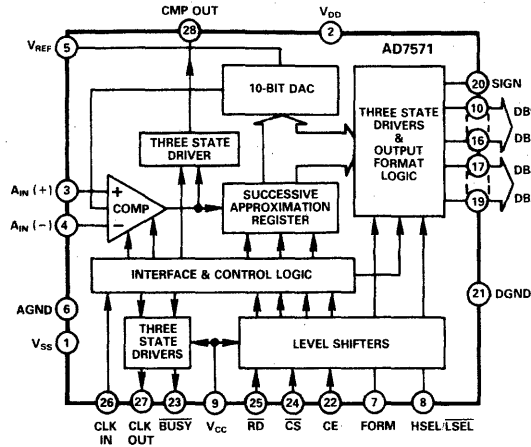
#### 28-PIN CERAMIC DIP (SUFFIX D)



#### 28-PIN PLASTIC DIP (SUFFIX N)



### AD7571 FUNCTIONAL BLOCK DIAGRAM



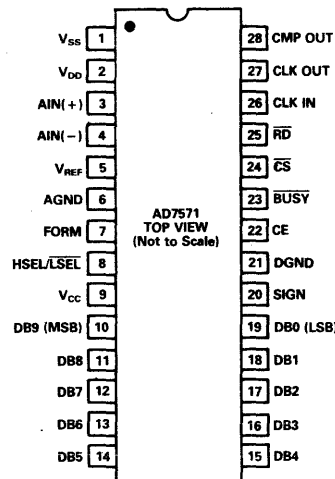
### ORDERING INFORMATION

Relative Accuracy (T <sub>min</sub> to T <sub>max</sub> )	Temperature Range and Package		
	Plastic 0 to +70°C	Cerdip <sup>1</sup> -25°C to +85°C	Side-Brazed Ceramic -55°C to +125°C
±1LSB	AD7571JN	AD7571AQ	AD7571SD
±1/2LSB	AD7571KN	AD7571BQ	AD7571TD

NOTE

<sup>1</sup>Analog Devices reserves the right to ship ceramic packages in lieu of cerdip packages.

### PIN CONFIGURATION



**AD7573**

**FEATURES**

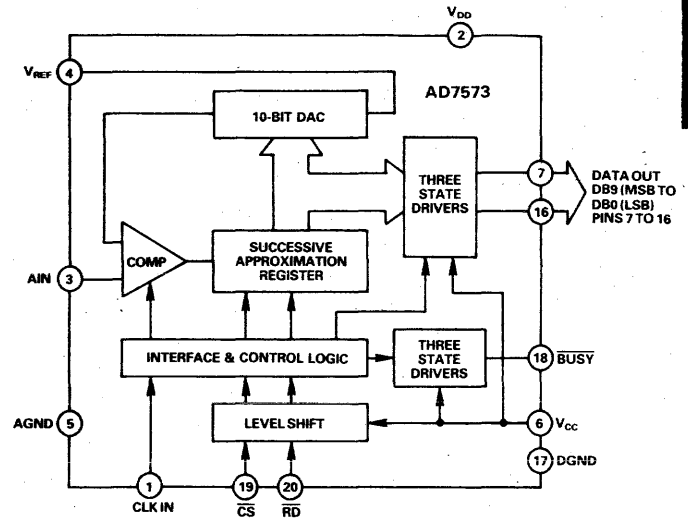
- 10-Bit Resolution
- No Missing Codes Over Full Temperature Range
- Conversion Time 80µs
- 0V to +10.24V Analog Input Voltage Range  
with Reference Voltage of +5.12V
- 10-Bit Parallel Data Output
- Easy Interface to Most Microprocessors
- Internal Clock Oscillator
- Single Supply Operation
- Monolithic

**GENERAL DESCRIPTION**

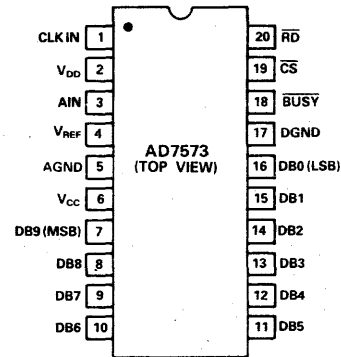
The AD7573 is a high speed, low cost 10-bit CMOS A/D converter which uses the successive approximation technique to provide a conversion time of 80µs. The device is designed for easy microprocessor interface using two standard control signals;  $\overline{CS}$  (decoded device address) and  $\overline{RD}$  ( $\overline{READ}$ / $\overline{WRITE}$  control), for conversion start and data read operations. Conversion results are available via a 10-bit wide three-state output bus.

The analog input voltage range is from 0V to +10.24V when using a reference voltage of +5.12V. Single +15V power supply operation is possible with all logic inputs and outputs being CMOS compatible. Using +5V/+15V supplies allows control inputs and data outputs to be TTL/CMOS compatible. The device is packaged in a small 20-pin 0.3" wide DIP.

**AD7573 FUNCTIONAL BLOCK DIAGRAM**



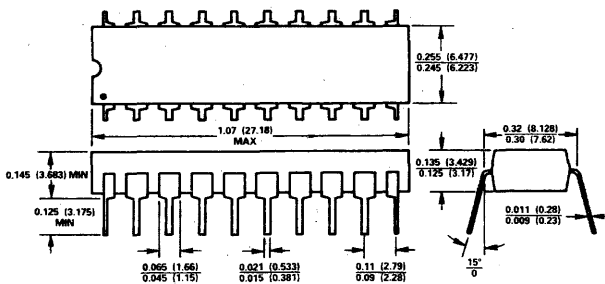
**PIN CONFIGURATION**



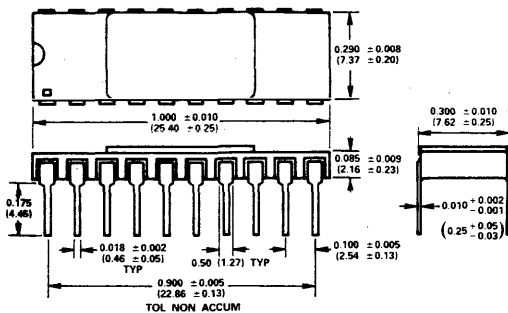
**MECHANICAL INFORMATION  
OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).

**20-PIN PLASTIC DIP (SUFFIX N)**



**20-PIN CERAMIC (SUFFIX D)**



**ORDERING INFORMATION**

Relative Accuracy ( $T_{min}$ to $T_{max}$ )	Temperature Range and Package		
	Plastic 0 to +70°C	Cerdip <sup>1</sup> -25°C to +85°C	Side-Braced Ceramic -55°C to +125°C
± 1LSB	AD7573JN	AD7573AQ	AD7573SD
± 1/2LSB	AD7573KN	AD7573BQ	AD7573TD

NOTE:  
<sup>1</sup>Analog Devices reserves the right to ship ceramic packages in lieu of cerdip packages.



# Complete, High Resolution 16-Bit A/D Converter

## AD ADC71/AD ADC72

### FEATURES

- Complete 16-Bit Converter With Reference and Clock
- $\pm 0.003\%$  Maximum Nonlinearity
- No Missing Codes to 14 Bits Over Full Temp Range
- Fast Conversion -  $35\mu\text{s}$  (14 Bit)
- Short Cycle Capability
- Parallel or Serial Logic Outputs
- Low Power: 850mW Typical
- Industry Standard Pin Out

### PRODUCT DESCRIPTION

The AD ADC71 and AD ADC72 are high resolution 16-bit hybrid IC analog to digital converters including reference, clock, and laser-trimmed thin-film components. The package is a compact 32-pin bottom-brazed ceramic (AD ADC71) or hermetic ceramic (AD ADC72) DIP. The thin-film scaling resistors allow analog input ranges of  $\pm 2.5\text{V}$ ,  $\pm 5\text{V}$ ,  $\pm 10\text{V}$ , 0 to  $+5\text{V}$ , 0 to  $+10\text{V}$ , and 0 to  $+20\text{V}$ .

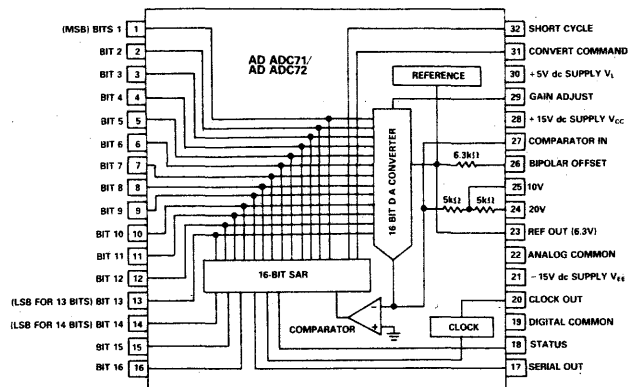
Important performance characteristics of the devices are maximum linearity error of  $\pm 0.003\%$  of FSR (AD ADC71K, AD ADC72K and B), and maximum conversion time of  $50\mu\text{s}$ . This performance is due to innovative design and the use of proprietary monolithic D/A converter chips. Laser-trimmed thin-film resistors provide the linearity and wide temperature range for no missing codes.

The AD ADC71 and AD ADC72 provide data in both parallel and serial form with corresponding clock and status output. All digital inputs and outputs are TTL compatible.

### PRODUCT HIGHLIGHTS

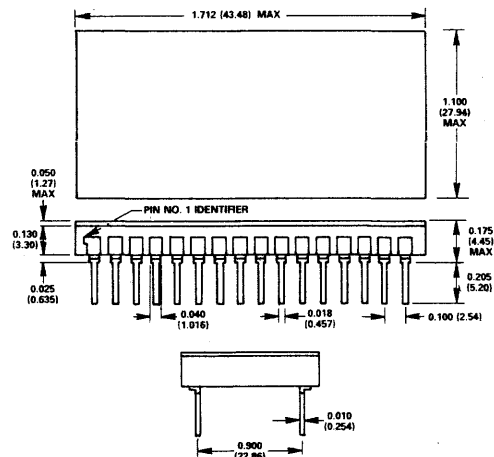
1. The AD ADC71 and AD ADC72 provide 16-bit resolution with maximum linearity error less than  $\pm 0.003\%$  ( $\pm 0.006\%$  for J and A grades) at  $25^\circ\text{C}$ .
2. Conversion time is  $45\mu\text{s}$  typical to 14 bits with short cycle capability.
3. The AD ADC72 (KD, BD) have no missing codes (to 14 bits) over the full specification temperature range.
4. Two binary codes are available on the AD ADC71 and AD ADC72 output. They are complementary straight binary (CSB) for unipolar input voltage ranges and complementary offset binary (COB) for bipolar input ranges. Complementary two's complement (CTC) coding may be obtained by inverting pin 1 (MSB).
5. The proprietary chips used in this hybrid design provide excellent stability over temperature and lower chip count for improved reliability.

### AD ADC71/AD ADC72 FUNCTIONAL BLOCK DIAGRAM



### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



### ORDERING GUIDE

Model	Linearity Error (Max)	Specification Temp Range	Package
AD ADC71JW	$\pm 0.006\%$ of FSR	0 to $+70^\circ\text{C}$	Ceramic
AD ADC71KW	$\pm 0.003\%$ of FSR	0 to $+70^\circ\text{C}$	Ceramic
AD ADC72JD	$\pm 0.006\%$ of FSR	0 to $+70^\circ\text{C}$	Hermetic
AD ADC72KD	$\pm 0.003\%$ of FSR	0 to $+70^\circ\text{C}$	Hermetic
AD ADC72AD	$\pm 0.006\%$ of FSR	$-25^\circ\text{C}$ to $+85^\circ\text{C}$	Hermetic
AD ADC72BD	$\pm 0.003\%$ of FSR	$-25^\circ\text{C}$ to $+85^\circ\text{C}$	Hermetic

### FEATURES

- 14-Bit Resolution
- 125kHz Word Rates
- Internal Track-and-Hold
- 40-Pin DIP

### APPLICATIONS

- FDM/TDM Transmultiplexers
- CAT/NMR Scanners
- PCM Systems
- Digital Audio

### GENERAL DESCRIPTION

The HAS-1409KM and HAS-1409AKM hybrid A/D converters offer designers performance characteristics which have never before been available.

Now, for the first time, high resolution and high speed come together in a hybrid package which includes an internal track-and-hold. The HAS-1409 units have resolutions of 14 bits, are capable of word rates up to 125kHz, and are complete with track-and-hold; all of these features are housed in a single 40-pin DIP package which dissipates only two watts.

The HAS-1409KM includes an internal clock, which allows the converter to be operated at any word rate from dc through 120kHz; the HAS-1409AKM is designed for applications which use an external system clock whose frequency establishes the user's optimum word rate, up to 125kHz.

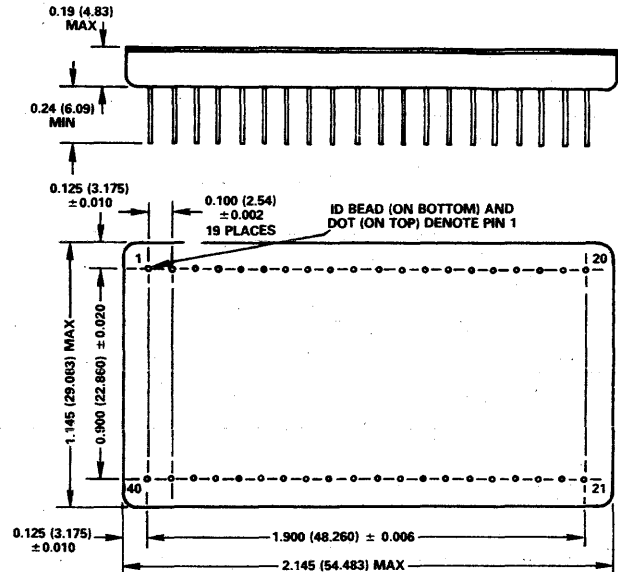
The HAS-1409 A/D has been characterized with a companion D/A converter, the HDD-1409KM, to emphasize the superior ac performance needed for use in Frequency Division Multiplex/Time Division Multiplex (FDM/TDM) transmultiplexer systems. Although specifically designed for these kinds of applications, it can also be used for other digital signal processing such as Computer Aided Tomography (CAT) and Nuclear Magnetic Resonance (NMR) scanners, and Pulse Code Modulation (PCM).

Conventional data converters often display errors at midscale which make them inadequate for use in the types of systems cited above. The unique Digitally Corrected Subranging technique pioneered by Analog Devices, used with other proprietary techniques, virtually cancels midscale errors in the HAS-1409, thereby eliminating a major source of system errors.

The logic outputs are TTL-compatible and are presented as 14 bits of parallel data. Buffer output registers and a 3-state format provide dual advantages of good drive and bus compatibility.

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



### HAS-1409 PIN DESIGNATION

PIN	FUNCTION	PIN	FUNCTION
1	± 10V INPUT	21	DIGITAL GROUND
2	± 5V INPUT	22	BIT 1 (MSB)
3	ANALOG GROUND	23	BIT 1 (MSB)
4	DIGITAL GROUND	24	BIT 2
5	+5V	25	BIT 3
6	+5V	26	BIT 4
7	N/C	27	BIT 5
8	N/C	28	BIT 6
9	+5V	29	ENABLE HIGH (MSBs)
10	DIGITAL GROUND	30	CLOCK/ENCODE
11	ENABLE LOW (LSBs)	31	ENCODE
12	BIT 14 (LSB)	32	+5V
13	BIT 13	33	DIGITAL GROUND
14	BIT 12	34	-15V
15	BIT 11	35	+15V
16	BIT 10	36	DIGITAL GROUND
17	BIT 9	37	ANALOG GROUND
18	BIT 8	38	ANALOG GROUND
19	BIT 7	39	+5V
20	DIGITAL GROUND	40	ANALOG GROUND

PIN 30 USED FOR CLOCK INPUT ON HAS-1409AKM; USED FOR ENCODE INPUT ON HAS-1409KM.  
 ALL +5V PINS ARE CONNECTED TOGETHER INTERNALLY (5, 6, 9, 32, 39). MUST ALSO BE CONNECTED TOGETHER EXTERNALLY CLOSE TO CASE.  
 ALL ANALOG GROUND PINS ARE CONNECTED TOGETHER INTERNALLY (3, 37, 38, 40).  
 ALL DIGITAL GROUND PINS ARE CONNECTED TOGETHER INTERNALLY (4, 10, 20, 21, 33, 36).  
 FOR BEST PERFORMANCE, ANALOG GROUND AND DIGITAL GROUND PINS MUST ALL BE CONNECTED TOGETHER AND TO GROUND EXTERNALLY AS CLOSE TO THE CASE AS POSSIBLE.

## HIGH PERFORMANCE/HIGH RESOLUTION FEATURE SELECTION CHART

		GENERAL PURPOSE				μP BUS COMPATIBLE				MULTI-CHAN	HIGH PERFORMANCE		μP BUS COMPATIBLE		HIGH RESOLUTION			
		AD570	AD571	AD573	AD ADC80	AD673	AD573	AD577	AD574	AD581	AD590	AD510	AD572	AD574	AD574A	AD575	AD ADC71	AD ADC72
Resolution	8 Bits	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
	10 Bits																	
	10 Bits & Sign																	
	12 Bits				•			•										
	13 Bits																	
	14 Bits																	
Conversion Time	1800ms																	
	50ms																	
	100μs																	
	50μs																	
	25μs	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
	15μs																	
Internal Reference		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Ratiometric Capability		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Low Power																		
Second Source					•													
Logic Compatibility	TTL	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
	CMOS																	
Operating Temperature Ranges	C = 0 to +70°C	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
	I = -25°C to +85°C	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
	M = -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
Dice Availability		•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•

NOTES  
<sup>1</sup> 80μs conversion time  
<sup>2</sup> 66.6μs/channel

### GENERAL PURPOSE SPECIFICATIONS (maximum @ +25°C unless otherwise noted)

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Conversion Time	Logic	Temp Range <sup>1</sup>	Package
AD570JD	8 Bits	±1/2LSB	Note 2	176	40μs	TTL	C	DIP
AD570SD	8 Bits	±1/2LSB	Note 2	80	40μs	TTL	M	DIP
AD571JD(SD)	10 Bits	±1/2LSB	Note 3	88/50/50	40μs	TTL & C/M/M	C/M/M	DIP
AD571KD	10 Bits	±1/2LSB	Note 2	44	40μs	CMOS	C	DIP
AD ADC80-10(Z)	10 Bits	±1/2LSB	±1/2LSB	30	21μs	TTL	I	DIP
AD ADC80-12(Z)	12 Bits	±1/2LSB	±1/2LSB	30	25μs	TTL	I	DIP
<b>μP COMPATIBLE</b>								
AD673J	8 Bits	±1/2LSB	Note 2	2 <sup>4</sup>	30μs	TTL	C	DIP
AD673S	8 Bits	±1/2LSB	Note 2	2 <sup>4</sup>	30μs	TTL	M	DIP
AD7574J(N)(AD)(SD)	8 Bits	±3/4LSB	±7/8LSB	Note 5	15μs	CMOS	C/I/M/M	DIP
AD7574KN(BD)(TD)	8 Bits	±1/2LSB	±3/4LSB	Note 5	15μs	CMOS	C/I/M/M	DIP
AD573J	10 Bits	±1LSB	Note 6	4 <sup>4</sup>	20μs	TTL	C	DIP
AD573K	10 Bits	±1/2LSB	Note 2	2 <sup>4</sup>	20μs	TTL	C	DIP
AD573S	10 Bits	±1LSB	Note 2	5 <sup>4</sup>	20μs	TTL	M	DIP
AD7571JN	10 Bits & Sign	±1LSB	Note 2	5 <sup>4</sup>	80μs	TTL & C	C	DIP
AD7571KN	10 Bits & Sign	±1/2LSB	Note 2	4 <sup>4</sup>	80μs	CMOS	C	DIP
AD7571AQ	10 Bits & Sign	±1LSB	Note 2	5 <sup>4</sup>	80μs	TTL & I	I	DIP
AD7571BQ	10 Bits & Sign	±1/2LSB	Note 2	4 <sup>4</sup>	80μs	CMOS	I	DIP
AD7571SD	10 Bits & Sign	±1LSB	Note 2	5 <sup>4</sup>	80μs	TTL & M	M	DIP
AD7571TD	10 Bits & Sign	±1/2LSB	Note 2	4 <sup>4</sup>	80μs	CMOS	M	DIP
AD574AJD	12 Bits	±1LSB	Note 7	9 <sup>4</sup>	35μs	TTL	C	DIP
AD574AKD	12 Bits	±1/2LSB	Note 2	5 <sup>4</sup>	35μs	TTL	C	DIP
AD574ALD	12 Bits	±1/2LSB	Note 2	2 <sup>4</sup>	35μs	TTL	C	DIP
AD574ASD	12 Bits	±1LSB	Note 7	20 <sup>4</sup>	35μs	TTL	M	DIP
AD574ATD	12 Bits	±1/2LSB	Note 2	10 <sup>4</sup>	35μs	TTL	M	DIP
AD574AUD	12 Bits	±1/2LSB	Note 2	5 <sup>4</sup>	35μs	TTL	M	DIP
AD7552	12 Bits & Sign	1 Count	Note 2	1 <sup>4</sup>	160 <sup>3</sup>	TTL & CMOS	C	DIP
<b>MULTI-CHANNEL</b>								
AD7581JN(AD) <sup>9</sup>	8 Bits	±1 7/8LSB	±1 7/8LSB	Note 5	66.6μs <sup>10</sup>	TTL & CMOS	C/I	DIP
AD7581KN(BD) <sup>9</sup>	8 Bits	±3/4LSB	±7/8LSB	Note 5	66.6μs <sup>10</sup>	CMOS	C/I	DIP
AD7581LN(CD) <sup>9</sup>	8 Bits	±1/2LSB	±3/4LSB	Note 5	66.6μs <sup>10</sup>	CMOS	C/I	DIP
<b>HIGH PERFORMANCE</b>								
AD5200BD Series	12 Bits	±1/2LSB	Note 2	Note 11	50μs	TTL	I	DIP
AD5200TD Series	12 Bits	±1/2LSB	Note 2	Note 11	50μs	TTL	M	DIP
AD574J(SD)	12 Bits	±1/2LSB	±1LSB	50	35μs	TTL	C/M/M	DIP
AD574K(LD)	12 Bits	±1/2LSB	Note 2	27/10	35μs	TTL	C	DIP
AD574TD(UD)	12 Bits	±1/2LSB	Note 2	25/12.5	35μs	TTL	M	DIP
AD572AD(BD)	12 Bit	±1/2LSB	±1/2LSB	30/15	25μs	TTL	I	DIP
AD572SD	12 Bits	±1/2LSB	±1/2LSB	15	25μs	TTL	M	DIP
AD5211B, AD5212B	12 Bits	±1/2LSB	Note 2	0.4% <sup>12</sup>	13μs	TTL	I	DIP
AD5211T, AD5212T	12 Bits	±1/2LSB	Note 2	0.4% <sup>12</sup>	13μs	TTL	M	DIP
AD5214B, AD5215B	12 Bits	±1/2LSB	Note 2	0.1% <sup>12</sup>	13μs	TTL	I	DIP
AD5214T, AD5215T	12 Bits	±1/2LSB	Note 2	0.1% <sup>12</sup>	13μs	TTL	M	DIP

NOTES  
<sup>1</sup> C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C  
<sup>2</sup> No missing codes over temperature  
<sup>3</sup> No missing codes  
<sup>4</sup> Gain error in LSB, T<sub>min</sub> to T<sub>max</sub>, max  
<sup>5</sup> ±2LSB max gain change from +25°C to T<sub>min</sub> or T<sub>max</sub>

<sup>6</sup> No missing codes over temperature, 9-bit resolution  
<sup>7</sup> No missing codes over temperature, 11-bit resolution  
<sup>8</sup> Milliseconds, typical  
<sup>9</sup> 8 channels

<sup>10</sup> Per channel  
<sup>11</sup> Absolute Accuracy = ±0.2%/0.1% of full scale input voltage over temperature range  
<sup>12</sup> Absolute accuracy error, T<sub>min</sub> to T<sub>max</sub>, max

### HIGH RESOLUTION SPECIFICATIONS (maximum @ +25°C unless otherwise noted)

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Conversion Time	Logic	Temp Range <sup>1</sup>	Package
AD7550BD	13 Bits	±1LSB		1	40ms typ	TTL & CMOS	I	DIP
AD7555KN(BD)	5 1/2 Digits	±10 Counts	Note 2	0.2	1760ms	CMOS	C/I	DIP
	4 1/2 Digits	±1 Count	Note 3	0.2	610ms	CMOS		
AD ADC71JW	16 Bits	±0.006 <sup>2</sup>	Note 6	15	50µs	DTL & TTL	C	DIP
AD ADC71KW	16 Bits	±0.003 <sup>3</sup>	Note 7	15	50µs	DTL & TTL	C	DIP
AD ADC72J(AD)	16 Bits	±0.006 <sup>2</sup>	Note 6	20(15)	50µs	DTL & TTL	C(I)	DIP
AD ADC72KD(BD)	16 Bits	±0.003 <sup>3</sup>	Note 7	20(15)	50µs	DTL & TTL	C(I)	DIP

NOTES  
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C  
<sup>2</sup>Count uncertainty due to noise: ±2 counts  
<sup>3</sup>Count uncertainty due to noise: ±1/2 count  
<sup>4</sup>No missing codes  
<sup>5</sup>% of FSR  
<sup>6</sup>No missing codes (to 13 bits) over temperature  
<sup>7</sup>No missing codes (to 14 bits) over temperature  
 (+10°C to +40°C - KW, 0 to +70°C - KD, -25°C to +85°C - BD)

### HIGH SPEED/VIDEO FEATURE SELECTION CHART

		HIGH SPEED											
		AD579	AD578	AD-ADC84/83 <sup>1</sup>	AD5740 <sup>1</sup>	HAS-0802	HAS-1002	HAS-1202	MAH-0801	MAH-1001	HAS-1202	AD6020	AD5010
Resolution	6 Bits												
	8 Bits												
	9 Bits												
	10 Bits	•	•	•	•	•	•	•	•	•	•	•	•
	12 Bits												
Conversion Time	<8µs												
	<3µs	•	•	•	•	•	•	•	•	•	•	•	•
	<1.1µs												
Word Rate	>5MHz												
	>16MHz												
	>50MHz												
	>100MHz												
Internal Reference		•	•	•	•	•	•	•	•	•	•	•	
Logic Compatibility	TTL	•	•	•	•	•	•	•	•	•	•	•	•
	ECL												
Operating Temperature Ranges	C = 0 to +70°C	•	•	•	•	•	•	•	•	•	•	•	•
	I = -25°C to +85°C	•	•	•	•	•	•	•	•	•	•	•	•
	M = -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•

NOTES  
<sup>1</sup>Second Source  
<sup>2</sup>Complete with Track and Hold.

### HIGH SPEED SPECIFICATIONS (maximum @ +25°C unless otherwise noted)

Model	Resolution	Relative Accuracy	Differential Nonlinearity	Gain T.C. ppm/°C	Conversion Time, min	Logic	Temp Range <sup>1</sup>	Package
AD578JN(D)	12 Bits	±1/2LSB	Note 2	30	6µs	TTL	C	DIP
AD578KN(KD)	12 Bits	±1/2LSB	Note 2	30	4.5µs	TTL	C	DIP
AD578LN(LD)	12 Bits	±1/2LSB	Note 2	30	3µs	TTL	C	DIP
AD578SD(SD/883B)	12 Bits	±1/2LSB	Note 2	50	6µs	TTL	M	DIP
AD578TD(TD/883B)	12 Bits	±1/2LSB	Note 2	30	4.5µs	TTL	M	DIP
AD5240KD	12 Bits	±1/2LSB	Note 2	30	5	TTL	C	DIP
AD5240SD	12 Bits	±1/2LSB	Note 2	25	5	TTL	M	DIP
AD ADC84-10(84-12)	10/12 Bits	±1/2LSB	Note 2	30/30	6/10	TTL	C	DIP
AD ADC85C-10(85C-12)	10/12 Bits	±1/2LSB	Notes 2 & 3	40/25	6/10	TTL	C	DIP
AD ADC85-10(85-12)	10/12 Bits	±1/2LSB	Notes 2 & 3	20/15	6/10	TTL	I	DIP
AD ADC85S-10(85S-12)	10/12 Bits	±1/2LSB	Note 2	25	8/10	TTL	M	DIP
AD579JN	10 Bits	±1/2LSB	Note 2	30	2.2	TTL	C	DIP
AD579KN	10 Bits	±1/2LSB	Note 2	30	1.8	TTL	C	DIP
AD579BD	10 Bits	±1/2LSB	Note 2	30	1.8	TTL	I	DIP
AD579TD	10 Bits	±1/2LSB	Note 2	30	1.8	TTL	M	DIP
HAS-0802	8 Bits	±1/4LSB	Note 2	30 typ	1.2µs typ	TTL	C <sup>4</sup>	DIP
HAS-1002	10 Bits	±1/2LSB	Note 2	30 typ	1.4µs typ	TTL	C <sup>4</sup>	DIP
HAS-1202	12 Bits	±1/2LSB	Note 2	30 typ	2.2µs typ	TTL	C <sup>4</sup>	DIP

NOTES  
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C  
<sup>2</sup>No missing codes  
<sup>3</sup>No missing codes over temperature  
<sup>4</sup>Extended temperature ranges available

### IC ADC FLASH ENCODERS SPECIFICATIONS (typical @ +25°C unless otherwise noted)

Model	Resolution	Relative Accuracy %	Aperture Uncertainty ps max	Word Rate	Output Code	Temperature Range °C	Size & Package Inches	Voltage Vols	Input Impedance
AD6020KD	6 Bits	±1/8LSB	20	dc to 50MHz	BIN	0 to +70	16-Pin DIP	±2.5 max	25pF <sup>1</sup>
AD5010KD	6 Bits	±1/8LSB	20	dc to 100MHz	BIN	0 to +70	16-Pin DIP	±2.5 max	25pF <sup>1</sup>

NOTE  
<sup>1</sup>Input Capacitance



# Voltage-to-Frequency and Frequency-to-Voltage Converter

## AD650

### FEATURES

- V/F Conversion to 1MHz
- Reliable Monolithic Construction
- Very Low Nonlinearity
  - 0.002% typ at 10kHz
  - 0.005% typ at 100kHz
  - 0.07% typ at 1MHz
- Input Offset Trimmable to Zero
- CMOS or TTL Compatible
- Unipolar, Bipolar, or Differential V/F
- V/F or F/V Conversion

### PRODUCT DESCRIPTION

The AD650 V/F/V (voltage-to-frequency or frequency-to-voltage converter) provides a combination of high frequency operation and low nonlinearity previously unavailable in monolithic form. The inherent monotonicity of the V/F transfer function makes the AD650 useful as a high-resolution analog-to-digital converter. A flexible input configuration allows a wide variety of input voltage and current formats to be used, and an open-collector output with separate digital ground allows simple interfacing to either standard logic families or opto-couplers.

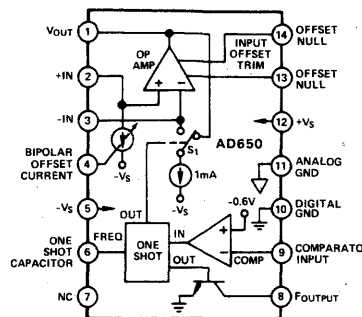
The linearity error of the AD650 is typically 20ppm (0.002% of full scale) and 50ppm (0.005%) maximum at 10kHz full scale. This corresponds to approximately 14-bit linearity in an analog-to-digital converter circuit. Higher full-scale frequencies or longer count intervals can be used for higher resolution conversions. The AD650 has a useful dynamic range of six decades, allowing extremely high resolution measurements. Even at 1MHz full scale, linearity is guaranteed less than 1000ppm (0.1%) on the AD650KN, BD and SD grades.

In addition to analog-to-digital conversion, the AD650 can be used in isolated analog signal transmission applications, phased-locked-loop circuits, and precision stepper motor speed controllers. In the F/V mode, the AD650 can be used in precision tachometer and FM demodulator circuits.

The input signal range and full-scale output frequency are user-programmable with two external capacitors and one resistor. Input offset voltage can be trimmed to zero with an external potentiometer.

The AD650JN and AD650KN are offered in a plastic 14-pin DIP package and are specified for the commercial (0 to +70°C) temperature range. For industrial temperature range (-25°C to +85°C) applications, the AD650AD and AD650BD are offered in a ceramic package. The AD650SD is specified for the full -55°C to +125°C extended temperature range.

### AD650 FUNCTIONAL BLOCK DIAGRAM

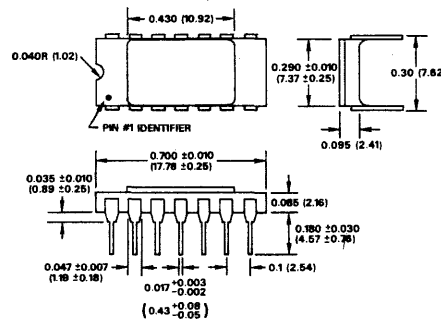


### ORDERING GUIDE

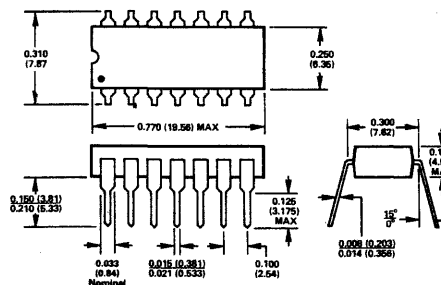
Part Number	Gain	1MHz Linearity	Specified Temperature Range °C	Package
	Tempco ppm/°C 100kHz			
AD650JN	150 typ	0.1% typ	0 to +70	Plastic DIP
AD650KN	150 typ	0.1% max	0 to +70	Plastic DIP
AD650AD	150 max	0.1% typ	-25 to +85	Ceramic
AD650BD	150 max	0.1% max	-25 to +85	Ceramic
AD650SD	150 max	0.1% max	-55 to +125	Ceramic

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



### "D" PACKAGE (TO-116) CERAMIC DIP



### "N" PACKAGE PLASTIC DIP

**V/F CONVERTERS**
**SPECIFICATIONS** (typical @ +25°C and  $V_S = \pm 15V$  dc unless otherwise noted)

	High Performance 500kHz ADVFC32K(B)(S)	Wide Band 1MHz AD650J(K)(S)	Economy/Versatile 0 to 150kHz AD537JH(KH)(SH)(JD)(KD)(SD)
<b>Model</b>			
<b>Analog Input</b>			
Voltage Signal Range - V min	0 to -10V	0 to -10V	- $V_S$ to (+ $V_S$ -4V) min
Current Signal Range - mA min	0 to +0.25	0 to +0.25	0.1 $\mu$ A to 1000 $\mu$ A
Overrange - % min	-	-	100
<b>Accuracy</b>			@ F = 10kHz
Nonlinearity, max - at rate F.S. range			
Voltage Input - %	±0.2	±0.5(±0.1)(±0.2)	0.15(0.07)(0.07)(0.15)(0.07)(0.07)
Current Input - %	±0.2	±0.5(±0.1)(±0.2)	±150(±50)(±150)(±150)(±50)(±150)
Gain vs. temperature, ppm/°C, max	150(100)(150)	150(100)(150)	±5(±2)(±2)(±5)(±2)(±2) max
Input Offset Voltage - mV vs. Temperature (0 to +70°C) - $\mu$ V/°C	±4 max ±30 max	±4 max ±30 max	±5(±1)(±10max) <sup>1</sup> (±5)(±1)(±10 max) <sup>1</sup>
<b>Response</b>			
Settling Time - $\mu$ s	1	1	-
Overload Recovery - ms	1 $\mu$ s	1 $\mu$ s	-
<b>Output</b>			
Waveform	← TRAIN OF TTL/DTL COMPATIBLE PULSES →		Symmetrical
Pulse Width - $\mu$ s	0.1 to 0.15/f max	0.1 to 0.15/f max	Square Wave, TTL, DTL, CMOS,
Pulse Polarity	← POSITIVE →		ECL, HN1L Compatible by Selection
Logic "1" (HIGH) Level - V min	Compatible by Resistor Selection	Compatible by Resistor Selection	of External Resistor
Logic "0" (LOW) Level - V max			
<b>Power Supply</b>			
Voltage, Rated Performance - V dc	±9 to ±18	±9 to ±18	+5 to +36 or ±5 to ±18
Current, Quiescent - mA	6	6	1.2
<b>Temperature Range<sup>2</sup></b>			
Rated Performance - °C	C(I)(M)	C(C)(M)	C(C)(M)(C)(C)(M)
<b>Case Size - Inches</b>	D-14-Pin DIP H-TO-100 Can	D-14-Pin DIP	D-14-Pin DIP; H-TO-100 Can

**NOTES**
<sup>1</sup> Guaranteed over -55°C to +125°C temperature range  
<sup>2</sup> C = 0 to +70°C, M = -55°C to +125°C

**MULTIPLEXERS AND SWITCHES**
**SPECIFICATIONS**
 $(T_A = +25^\circ\text{C}, V_{\text{SUPPLY}} = \pm 15V$  unless otherwise noted)

Type <sup>1</sup>	Function	$R_{\text{ON}}$ $\Omega$ , max	Off Leakage nA, max	Temp Range <sup>2</sup>	Logic Com- patibility
AD7501JN(JD)		10	10	C/I	CMOS
AD7501KN(KD)	8-Channel	300	10	C/I	TTL/CMOS
AD7501SD			5	M	TTL/CMOS
AD7502JN(JD)	Dual		5	C/I	CMOS
AD7502KN(KD)	4-Channel	300	5	C/I	TTL/CMOS
AD7502SD	(differential)		3	M	TTL/CMOS
AD7503JN(JD)		10	10	C/I	CMOS
AD7503KN(KD)	8-Channel	300	10	C/I	TTL/CMOS
AD7503SD			5	M	TTL/CMOS
AD7506JN(JD)		450	20	C/I	CMOS
AD7506KN(KD)	16-Channel	450	20	C/I	TTL/CMOS
AD7506SD		400	10	M	CMOS
AD7506TD		400	10	M	TTL/CMOS
AD7507JN(JD)	Dual	450	10	C/I	CMOS
AD7507KN(KD)	8-Channel	450	10	C/I	TTL/CMOS
AD7507SD	(differential)	400	5	M	CMOS
AD7507TD		400	5	M	TTL/CMOS

**NOTES**
<sup>1</sup> Suffix "N": plastic DIP; Suffix "D": ceramic DIP

<sup>2</sup> C: Commercial (0 to +70°C)

<sup>3</sup> I: Industrial (-25°C to +85°C)

<sup>4</sup> M: Military (-55°C to +125°C)

**SPECIFICATIONS**
 $(T_A = +25^\circ\text{C}, V_{\text{SUPPLY}} = \pm 15V$  unless otherwise noted)

Type <sup>1</sup>	Function	$R_{\text{ON}}$ $\Omega$ , max	OFF Leakage	Temp Range <sup>2</sup>	Logic Com- patibility
ADG200CJ		80	5nA, max	C	TTL/CMOS
ADG200BP		80	5nA, max	I	TTL/CMOS
ADG200BA	Dual SPST	80	5nA, max	I	TTL/CMOS
ADG200AP		70	2nA, max	M	TTL/CMOS
ADG200AA		70	2nA, max	M	TTL/CMOS
ADG201CJ		100	5nA, max	C	TTL/CMOS
ADG201BP	Quad SPST	100	5nA, max	I	TTL/CMOS
ADG201AP		80	1nA, max	M	TTL/CMOS
AD7510D1JN(JD)	Quad SPST		5nA, max	C/I	CMOS
AD7510D1KN(KD)	Note 3	100	5nA, max	C/I	TTL/CMOS
AD7510D1SD			3nA, max	M	TTL/CMOS
AD7511D1JN(JD)	Quad SPST		5nA, max	C/I	CMOS
AD7511D1KN(KD)	Note 3	100	5nA, max	C/I	TTL/CMOS
AD7511D1SD			3nA, max	M	CMOS
AD7511D1DIT			3nA, max	M	TTL/CMOS
AD7512D1JN(JD)	Dual SPST		15nA, max	C/I	CMOS
AD7512D1KN(KD)	Note 3	100	15nA, max	C/I	TTL/CMOS
AD7512D1SD			9nA, max	M	CMOS
AD7512D1DIT			9nA, max	M	TTL/CMOS
AD7590D1KN	Quad SPST	90	5nA, max	C	TTL/CMOS
AD7590D1BD	with Data	90	5nA, max	I	TTL/CMOS
AD7591D1KN	Latches	90	5nA, max	C	TTL/CMOS
AD7591D1BD		90	5nA, max	I	TTL/CMOS
AD7592D1KN	Dual SPST	90	5nA, max	C	TTL/CMOS
AD7592D1BD	with Data Latches	90	5nA, max	I	TTL/CMOS

**NOTES**
<sup>1</sup> Suffixes: N - Plastic DIP

P, D - Ceramic DIP

A, H - TO-100

<sup>2</sup> C: Commercial (0 to +70°C)

<sup>3</sup> I: Industrial (-25°C to +85°C)

<sup>4</sup> M: Military (-55°C to +125°C)

<sup>5</sup> Dielectrically isolated features latchup-free overvoltage proof operation.

<sup>6</sup>  $V_{\text{DD}} - V_{\text{SS}}$  (supply voltages) = +15V

<sup>7</sup>  $V_{\text{DD}}$  (supply voltage) = 8V

## INTEGRATED CIRCUIT REFERENCES

The accuracy of all measurements is determined by the reference employed. Analog Devices' references use buffered low TC precision zener diodes or patented laser wafer trimmed monolithic bandgap circuitry. Both include output operational amplifiers for optimum load regulation.

The AD2720 precision hybrid reference provides the highest initial accuracy and lowest temperature coefficient. The proprietary temperature compensated design and active laser trimming result in ultra high precision performance previously available only in oven-regulated modules.

**AD584  $V_{OUT} = 10.000V$  or  $7.500V$  or  $5.000V$  or  $2.500V$**

The AD584 is a unique bandgap reference. Depending on user requirements it may be connected to provide any one of four standard voltages without additional external parts.

## SPECIFICATIONS

(typical @  $V_{IN} = +15V$  and  $+25^{\circ}C$  unless otherwise noted)

Model	Maximum Error in mV for Nominal Output of:				Temp Stability ppm/ $^{\circ}C$ max	Temp Range <sup>1</sup>
	2.5V	5V	7.5V	10V		
AD584JH	$\pm 7.5$	$\pm 15$	$\pm 22$	$\pm 30$	30	C
AD584KH	$\pm 3.5$	$\pm 6$	$\pm 8$	$\pm 10$	15	C
AD584LH	$\pm 2.5$	$\pm 3$	$\pm 4$	$\pm 5$	5	C
AD584SH	$\pm 7.5$	$\pm 15$	$\pm 22$	$\pm 30$	30	M
AD584TH	$\pm 3.5$	$\pm 6$	$\pm 8$	$\pm 10$	15	M

NOTE  
<sup>1</sup> C = 0 to  $+70^{\circ}C$ , M =  $-55^{\circ}C$  to  $+125^{\circ}C$

## SPECIFICATIONS

(min-max @  $E_{IN} = +15V$  or  $-15V$  and  $+25^{\circ}C$  unless otherwise noted)

Model	Output Voltage Volts	Output Voltage Tolerance % Error	Output Current mA	Temp Stability ppm/ $^{\circ}C$	Time Stability (typ)	Temp Range <sup>1</sup>	Model	Output Voltage Volts	Output Voltage Tolerance % Error	Output Current mA	Temp Stability ppm/ $^{\circ}C$	Time Stability (typ)	Temp Range <sup>1</sup>	
AD580JH(KH)(LH)(MH)	2.5	3.0(1.0)(0.4)(0.4)	10	85(40)(25)(10)	25 $\mu$ V/Month 25 $\mu$ V Long Term	C	AD2700(S) AD2700(L)(U)	10.000 10.000	$\pm 0.05$ $\pm 0.025$	10 10	10(3) 3	100ppm per 1000 Hours @ $+55^{\circ}C$	I/M	
AD580SH(TH)(UH)	2.5	1.0(0.4)(0.4)	10	55(25)(10)	25 $\mu$ V/Month 25 $\mu$ V Long Term	M	AD2701(S) AD2701(L)(U)	-10.000 -10.000	$\pm 0.05$ $\pm 0.025$	10 10	10(3) 3	100ppm per 1000 Hours @ $+55^{\circ}C$	I/M	
AD581JH(KH)(LH)	10.000	$\pm 0.3(0.1)(0.05)$	10	30(15)(5)	25ppm per 1000 Hours Noncumulative	C	AD2702(S) AD2702(L)(U)	$\pm 10.000$ $\pm 10.000$	$\pm 0.05$ $\pm 0.025$	10 10	10(5) 5(3)	100ppm per 1000 Hours @ $+55^{\circ}C$	I/M	
AD581SH(TH)(UH)	10.000	$\pm 0.3(0.1)(0.05)$	10	30(15)(5)	25ppm per 1000 Hours Noncumulative	M	AD2710KN(LN) AD2712KN(LN)	+10.000 $\pm 10.000$	$\pm 0.01$ $\pm 0.01$	10 10	2(1) 2(1)		C	
AD589JH(KH)(LH)(MH)	1.235	-2.8, +1.2	5	100(50)(25)(10)		C	AD2720AH AD2720BH AD2720CH AD2720SH AD2720TH	+10.000 +10.000 +10.000 +10.000 +10.000	$\pm 0.02$ $\pm 0.01$ $\pm 0.01$ $\pm 0.02$ $\pm 0.01$	5 5 5 5 5	4 2 1 4 2	25ppm per 1000 Hours @ $+25^{\circ}C$	I I I M M	
AD589SH(TH)(UH)	1.235 typ	-2.8, +1.2	5	100(50)(25)		M								

NOTES  
<sup>1</sup> C = 0 to  $+70^{\circ}C$ , I =  $-25^{\circ}C$  to  $+85^{\circ}C$ , M =  $-55^{\circ}C$  to  $+125^{\circ}C$

## FEATURE SELECTION CHART

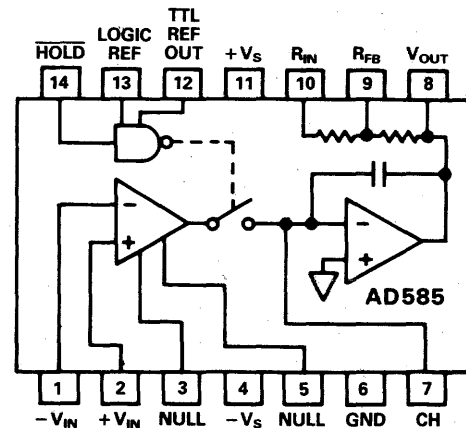
		AD589	AD580	AD1403	AD581	AD584	AD2700	AD2710	AD2720
Output Voltage Range	1.235V 2.5V 5.0V 7.5V +10.00V -10.00V $\pm 10.00V$	•	•	•	•	•	•	•	•
Output Voltage Tolerance	$\leq \pm 0.4\%$ $\leq \pm 0.05\%$ $\leq \pm 0.025\%$ $\leq \pm 0.012\%$		•	•	•	•	•	•	•
Temperature Stability	$\leq 25$ ppm/ $^{\circ}C$ $\leq 10$ ppm/ $^{\circ}C$ $\leq 5$ ppm/ $^{\circ}C$ $\leq 1$ ppm/ $^{\circ}C$	•	•	•	•	•	•	•	•
Temperature Range	0 to $+70^{\circ}C$ $-25^{\circ}C$ to $+85^{\circ}C$ $-55^{\circ}C$ to $+125^{\circ}C$	•	•	•	•	•	•	•	•
Package Style	Hermetic Package Plastic Package	•	•	•	•	•	•	•	•
Diets Available		•	•	•	•	•	•	•	•

**FEATURES**

- Fast 2.5 $\mu$ s Acquisition Time to  $\pm 0.01\%$
- Low Droop Rate: 0.5mV/ms
- Low Offset: 1mV
- Sample/Hold Offset Step: 1mV
- Aperture Jitter: 0.5ns
- Military Temperature Range:  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$
- Internal Hold Capacitor

**APPLICATIONS**

- Data Acquisition Systems
- Data Distribution Systems
- Analog Delay & Storage
- Peak Amplitude Measurements

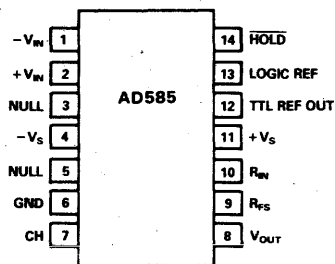
**AD585 FUNCTIONAL BLOCK DIAGRAM**

**PRODUCT DESCRIPTION**

The AD585 is a monolithic sample-and-hold circuit consisting of a high performance operational amplifier in series with an ultra-low leakage analog switch and a FET input integrating amplifier. An internal holding capacitor and connections to the internal feedback resistors, completes the sample and hold.

With the analog switch closed, the AD585 functions like a standard op amp; any feedback network may be connected around the device to control gain and frequency response. With the switch open, the capacitor holds the output at its previous level.

The AD585 offers performance previously unavailable in monolithic sample-and-hold amplifiers. The combination of a fast acquisition time (2.5 $\mu$ s to 0.01%) and low offset step (1mV) are suitable for high speed 12-bit data acquisition systems.

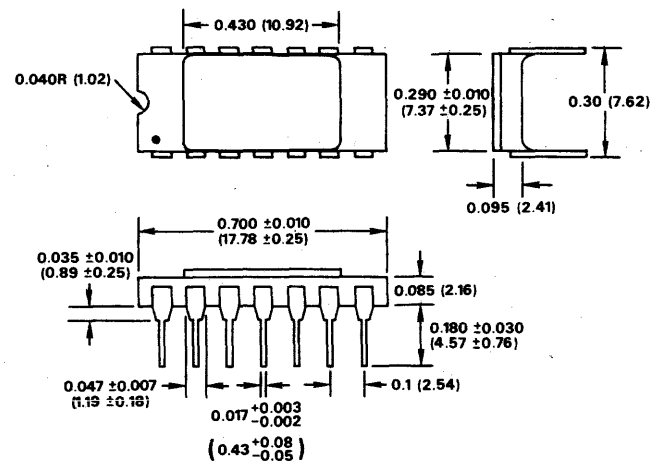
The device is available in two versions: the "J" specified for operation over the 0 to  $+70^{\circ}\text{C}$  commercial temperature range and the "S" specified over the full extended temperature range  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ .

**PIN CONFIGURATION**

**PRODUCT HIGHLIGHTS**

1. The droop rate is only 0.5mV/ms so that it may be used in slower high accuracy systems without the loss of accuracy.
2. The fast acquisition time and low aperture make it suitable for very high speed data acquisition systems.
3. The AD585 has internal pretrimmed application resistors for applications versatility.
4. The AD585 is complete with an internal hold capacitor for ease of use and capacitance can be added externally to achieve higher accuracy.

**OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).





### FEATURES

- Aperture Jitter of 5ps
- Acquisition Time 10ns
- Output Current  $\pm 40\text{mA}$
- Slew Rate  $300\text{V}/\mu\text{s}$

### APPLICATIONS

- Data Acquisition Systems
- Radar Systems
- Instrumentation Systems
- Medical Electronics

### GENERAL DESCRIPTION

The Analog Devices HTS-0010 Track-and-Hold is another example of Analog's continuing efforts to advance the state of the art in high-speed circuits.

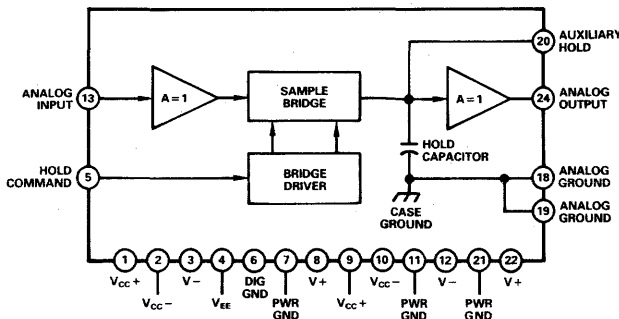
The HTS-0010 adds breadth to a line of devices which offers designers the industry's widest range of track-and-hold and sample-and-hold units.

Its pinouts are similar to its predecessor HTS-0025 Track-and-Hold, but it provides enhanced performance in many of the characteristics established by that device. Two pins which are unused on the HTS-0025 are used on the HTS-0010, but with those exceptions, the two devices have identical pin assignments. This plug-in compatibility gives designers remarkable flexibility in selecting those parameters which are optimum for their applications.

The HTS-0010 Track-and-Hold (T/H) uses many of the proven design concepts which have made the HTS-0025 T/H the standard of comparison for high-speed circuits of this type. A dc-coupled Schottky diode bridge is driven by a high impedance buffer amplifier and followed by a low impedance output amplifier to achieve the best possible combination of speed and drive capabilities.

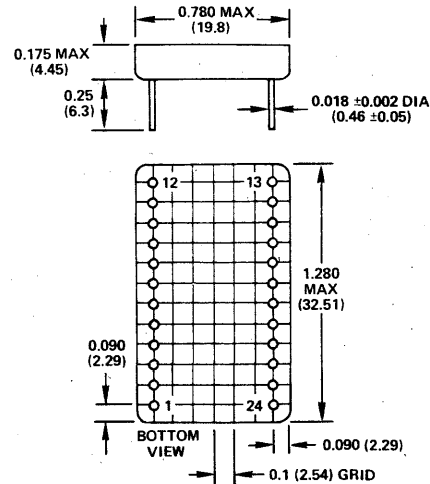
All models of the HTS-0010 are housed in a standard 24-pin metal DIP. The unit operating over a temperature range of 0 to  $+70^\circ\text{C}$  is HTS-0010KD; the unit for a range of  $-55^\circ\text{C}$  to  $+100^\circ\text{C}$  is HTS-0010SD.

### HTS-0100 FUNCTIONAL BLOCK DIAGRAM



### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



DOT ON TOP AND CERAMIC BEAD ON BOTTOM INDICATE POSITION OF PIN 1. PINS ARE GOLD PLATED.

### PIN DESIGNATIONS

PIN	FUNCTION
1	V <sub>cc</sub> + (+5V)
2	V <sub>cc</sub> - (-5V)
3	V <sub>-</sub> (-15V)
4	V <sub>ee</sub> (-5.2V)
5	HOLD COMMAND
6	DIGITAL GROUND
7	POWER GROUND
8	V <sub>+</sub> (+15V)
9	V <sub>cc</sub> + (+5V)
10	V <sub>cc</sub> - (-5V)
11	POWER GROUND
12	V <sub>-</sub> (-15V)
13	ANALOG INPUT
14	N/A
15	N/A
16	N/A
17	N/A
18	ANALOG GROUND
19	ANALOG GROUND
20	AUXILIARY HOLD
21	POWER GROUND
22	V <sub>+</sub> (+15V)
23	N/A
24	ANALOG OUTPUT

POWER GROUND (PINS 7, 11 AND 21), ANALOG GROUND (PINS 18 AND 19), AND DIGITAL GROUND (PIN 6) MUST BE CONNECTED TOGETHER AND TO A LOW-IMPEDANCE GROUND FOR PROPER OPERATION. MAKE CONNECTIONS AS CLOSE TO DEVICE AS POSSIBLE. HYBRID CASE IS CONNECTED TO ANALOG GROUND INTERNALLY.

**FEATURES**

700ns Acquisition Time  
 <750mW Power Dissipation  
 14-Pin DIP  
 0.01% Linearity

**APPLICATIONS**

Data Acquisition Systems  
 Data Distribution Systems  
 Analog Delay and Storage  
 Peak Amplitude Measurements

**GENERAL DESCRIPTION**

The Analog Devices HTC-0500 Track/Sample Hold is a remarkable combination of speed and low power dissipation in a 14-pin DIP. Its low cost makes it extremely attractive for a wide range of applications which were often uneconomical until now.

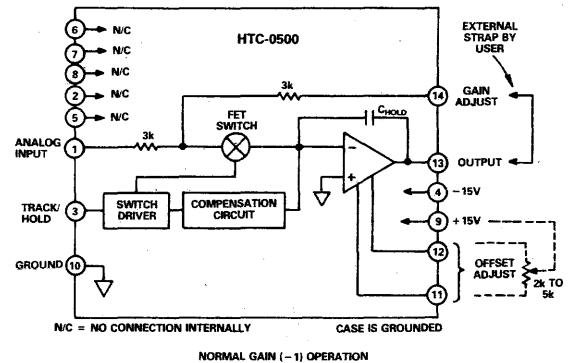
Exceptional speed and minimum power in a small, cost-effective package are not the only characteristics which make this unit worth serious consideration for a variety of uses. The innovative design ideas which have been included make it possible for the user to vary the gain of this inverting amplifier.

In many instances, Track/Sample-Hold devices may allow the user to decrease the gain, but increasing the gain is impossible. This is because the majority of these units close the feedback loop internally.

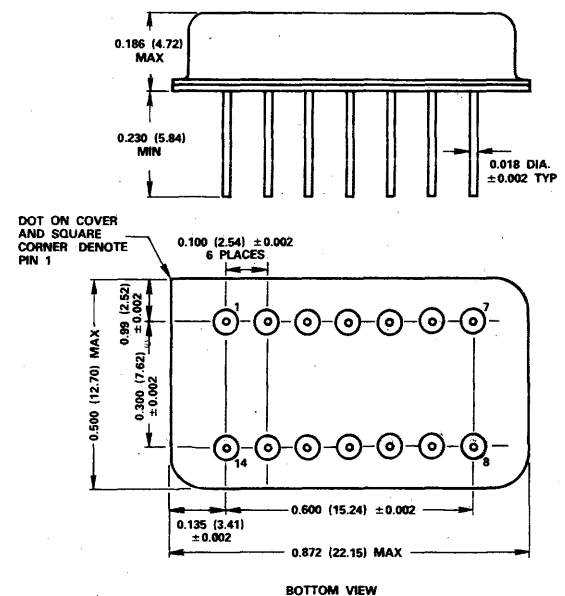
The HTC-0500, however, gives the designer flexibility when it is incorporated into its system application because it can be varied around its normal unity gain. In fact, as shown in Figure 1, the user must close the feedback loop externally with a strap to get proper operation.

The HTC-0500 is a perfect choice for use with Analog Devices' converters such as the HAS-1202, AD578, and AD579 in applications which do not require the speed of the model HTC-0300 Track-and-Hold, but require higher speed than the AD578-85.

The standard unit is housed in a metal dual in-line package; its model number is HTC-0500AM. A temperature range of  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  is available with model HTC-0500SM.

**HTC-0500 FUNCTIONAL BLOCK DIAGRAM**

**MECHANICAL DIMENSIONS**

Dimensions shown in inches and (mm).





# Fast, Complete, 16-Channel μP Compatible 12-Bit Data Acquisition System

## AD364

### ADVANCED TECHNICAL DATA

#### FEATURES

- Complete Data Acquisition System in 2-Package IC Form
- Full 8- or 16-Bit Microprocessor Bus Interface
- 16 Single-Ended or 8 Differential Channels with Switchable Mode Control
- True 12-Bit Operation: Nonlinearity  $\leq \pm 0.012\%$
- Guaranteed No Missing Codes Over Specified Temperature Range
- High Throughput Rate: 20kHz
- Fast Successive Approximation Conversion:  $25\mu s$
- Buried Zener Reference for Long-Term Stability and Low Gain TC
- Small Size: Requires Only 2.8 Square Inches
- Short-Cycle Capability
- Low Power: 1.4 Watts
- Extended Temperature Range:  $-55^\circ C$  to  $+125^\circ C$

#### PRODUCT DESCRIPTION

The AD364 is a complete 16 channel, microprocessor compatible, 12-bit data acquisition system in integrated circuit form. The AD364 design is implemented with linear compatible LSI chips, active laser trimming and hybrid technology resulting in maximum performance and flexibility.

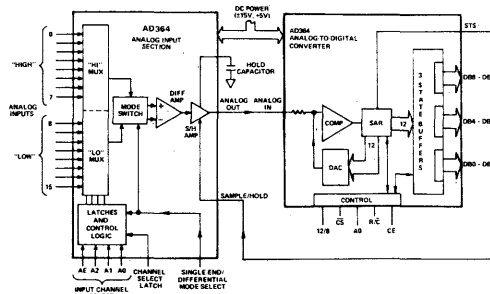
The AD364 consists of two separate functional blocks, each in a hermetically sealed dual-in-line package. The analog input section contains two eight-channel multiplexers, a differential amplifier, a sample-and-hold, a channel address register and control logic. The multiplexers may be connected to the differential amplifier in either an 8-channel differential or 16-channel single-ended configuration. A unique feature of the AD364 is an internal user-controllable analog switch that connects the multiplexers in either a single-ended or differential mode. This allows a single device to perform in either mode without hard-wire programming and permits a mixture of single-ended and differential sources to be interfaced to an AD364 by dynamically switching the input mode control.

The ADC section contains a complete 12-bit successive approximation ADC, including internal clock, precision 10 volt reference, comparator and bus interface. The ADC uses the newly-developed LCI (Linear-Compatible Integrated Injection Logic) process to provide the low power logic necessary to make a high speed 12-bit ADC and 3-state output buffer circuitry for direct interface to an 8-, 12- or 16-bit microprocessor bus.

#### AD364 ORDERING GUIDE

Model	Linearity	Temp. Range
AD364JD	$\pm 0.024\%$	0 to $+70^\circ C$
AD364KD	$\pm 0.012\%$	0 to $+70^\circ C$
AD364SD	$\pm 0.024\%$	$-55^\circ C$ to $+125^\circ C$
AD364TD	$\pm 0.024\%$	$-55^\circ C$ to $+125^\circ C$

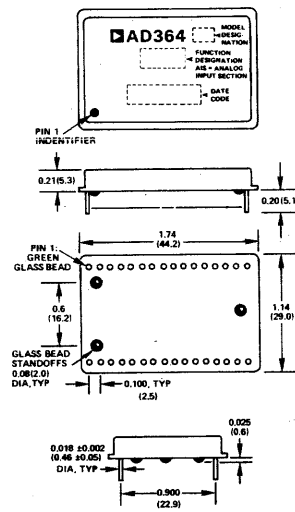
### AD364 FUNCTIONAL BLOCK DIAGRAM



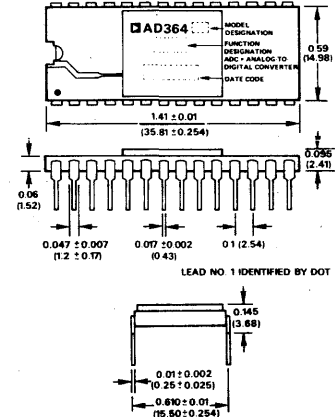
### OUTLINE DIMENSIONS PACKAGE SPECIFICATIONS

Dimensions shown in inches and (mm).

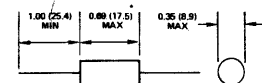
#### ANALOG INPUT SECTION



#### ANALOG-TO-DIGITAL CONVERTER

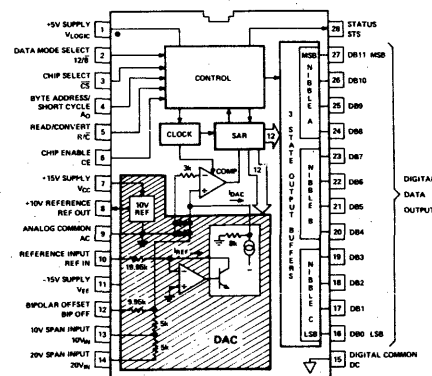


#### HOLD CAPACITOR



- NOTES:  
 1) PACKAGE: KOVAR WITH 100μIN. MIN. NICKEL PLATE  
 2) PINS: KOVAR WITH 50μIN. MIN. 24K GOLD PLATE  
 3) PACKAGE AND PINS MEET ALL REQUIREMENTS OF MIL-STD-883  
 4) TOLERANCES, UNLESS OTHERWISE NOTED:  
 a) XX: ±0.1 (25)  
 b) XXX: ±0.005 (1.27)

\*THIS DIMENSION IS FOR POLYSTYRENE CAPACITOR SUPPLIED WITH A AND K GRADES.  
 MAX BODY LENGTH OF TEFLOW CAPACITOR SUPPLIED WITH S AND T GRADES IS 1.00"



## IC SAMPLE-HOLD AMPLIFIERS

The AD582, AD583 and AD585 are monolithic sample-and-hold amplifiers with high performance internal amplifiers and a low leakage analog switch. An external hold capacitor is used with the AD582 and AD583, while the AD585 includes an on-chip hold capacitor.

The AD582 is the lowest cost SHA; yet it can replace many modular and hybrid units. For faster acquisition time or greater precision, the AD583 and AD585 are recommended.

Greater speed is provided by the HTC-0500, with an acquisition time of 700ns. The highest speed IC track-and-holds are the HTC-0300, HTC-0300A and HTS-0025. These units have acquisition times, respectively, of 170ns, 150ns and 30ns.

The HTC-0500, HTC-0300 and HTC-0300A are used with fast ADC's such as the HAS Series, the AD578, or AD579. The HTS-0025 is optimized for use with "video" ADC's and is also an excellent choice for use as a deglitcher at the output of high-speed current output DAC's. The AD582 is a second source for a generic industry-standard device.

## SPECIFICATIONS (typical @ +25°C unless otherwise noted)

Model	Open Loop Gain RL=2k $\Omega$ k min	Output V <sub>min</sub> /mA min	Gain Bandwidth MHz	Acquisition Time (0.1%) <sup>1</sup>	Aperture Time ns	Aperture Uncertainty	Droop Rate mV/ms	Oper. Temp <sup>2</sup>	Package
AD582K	25 <sup>3</sup>	$\pm 10/\pm 10$	1.5	6 $\mu$ s	150	15ns	100pA <sup>4</sup>	C	14-Pin DIP TO-100
AD582S	25 <sup>3</sup>	$\pm 10/\pm 10$	1.5	6 $\mu$ s	150	15ns	100pA <sup>4</sup>	M	14-Pin DIP TO-100
AD583K	25 <sup>3</sup>	$\pm 10/\pm 10$	2	4 $\mu$ s	50	5ns	50pA <sup>4</sup>	C	14-Pin DIP
AD585J	30 <sup>3</sup>	$\pm 10/\pm 10$	2.5	2.5 $\mu$ s <sup>5</sup>	10	1ns	0.5	C	14-Pin DIP
AD585S	30 <sup>3</sup>	$\pm 10/\pm 10$	2.5	2.5 $\mu$ s <sup>5</sup>	10	1ns	0.5	M	14-Pin DIP
ADSHC-85	+1	$\pm 10/\pm 10^6$	0.2	4.5 $\mu$ s <sup>5</sup>	25	0.5ns	0.2	C	14-Pin DIP
ADSHC-85ET	+1	$\pm 10/\pm 10^6$	0.2	4.5 $\mu$ s <sup>5</sup>	25	0.5ns	0.2	M	14-Pin DIP
HTS-0010KD/SD	0.96	$\pm 2/\pm 40$	60	15ns	-2	5ps	0.1	C/M	24-Pin DIP
HTS-0025	0.92	$\pm 2/\pm 50$	30	30ns	10	20ps	200	C	24-Pin DIP
HTS-0025M	0.92	$\pm 2/\pm 50$	20	30ns	10	20ps	200	MR	24-Pin DIP
HTC-0300	-1	$\pm 10/\pm 50$	8	170ns	10	100ps	5	C	24-Pin DIP
HTC-0300A	-1	$\pm 10/\pm 50$	10	150ns	6	100ps	1	C	24-Pin DIP
HTC-0300AM	-1	$\pm 10/\pm 50$	10	150ns	6	100ps	1	M	24-Pin DIP
HTC-0300M	-1	$\pm 10/\pm 50$	8	170ns	6	100ps	5	MR	24-Pin DIP
HTC-0500AM	-1	$\pm 12/\pm 15$	2	700ns	30	60ps	0.5	I	14-Pin DIP
HTC-0500SM	-1	$\pm 12/\pm 15$	2	700ns	30	60ps	0.5	M	14-Pin DIP

### NOTES

<sup>1</sup> A<sub>v</sub>=1, R<sub>L</sub>=2k, C<sub>L</sub>=50pF

<sup>2</sup> C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C, MR = -55°C to +100°C

<sup>3</sup> Open-loop gain - k min, R<sub>L</sub> = 2k $\Omega$

<sup>4</sup> Droop current; rate depends on external capacitance

<sup>5</sup> Acquisition Time to 0.01%

<sup>6</sup> Typical outputs



# DATA CONVERSION

ANALOG-TO-DIGITAL CONVERTERS							
Description	Model	Resolution (Bits)	Linearity Error (% of FSR) max	Conversion Time ( $\mu$ sec) max	Tempco Bipolar (ppm of FSR/ $^{\circ}$ C) max	Input Range (V)	Package
-55 $^{\circ}$ C to +200 $^{\circ}$ C	ADC10HT	12	$\pm 0.012$ to $\pm 0.048$	50	$\pm 34$ to $\pm 63$	$\pm 5$ , $\pm 10$ , $+10$ , $+20$	Hermetic DIP
High Speed	ADC60	8 - 12	$\pm 0.195$ to $\pm 0.0244$	0.88 - 3.5	$\pm 15$ to $\pm 20$	$\pm 2.5$ to $+20$	Module
High Resolution	ADC71	16	$\pm 0.003$ to $\pm 0.006$	50	$\pm 15$	$\pm 2.5$ to $\pm 20$	DIP
	ADC72	16	$\pm 0.003$ to $\pm 0.006$	50	$\pm 12$ to $\pm 17$	$\pm 2.5$ to $\pm 20$	Hermetic DIP
	ADC73/731	16	$\pm 0.0015$ to $\pm 0.00075$	170	$\pm 10$	$\pm 5$ , $\pm 10$ , $+10$ , $+20$	Module
High Speed, High Resolution	ADC76	16	$\pm 0.003$ to $\pm 0.006$	15	$\pm 15$	$\pm 2.5$ to $+20$	DIP
Low Cost	ADC80	10 - 12	$\pm 0.048$ to $\pm 0.012$	21 - 25	$\pm 23$	$\pm 2.5$ to $\pm 10$	DIP
High Speed, Low Cost	ADC82	8	$\pm 0.2$	2.8	$\pm 60$	$\pm 2.5$ to $+20$	DIP
	ADC84	10 - 12	$\pm 0.012$ to $\pm 0.048$	6 - 10	$\pm 21$	$\pm 2.5$ to $\pm 10$	Hermetic DIP
High Speed, Low Drift	ADC85	10 - 12	$\pm 0.012$ to $\pm 0.048$	6 - 10	$\pm 13$ to $\pm 26$	$\pm 2.5$ to $\pm 10$	Hermetic DIP
Ultra-High Speed	ADC803	12	$\pm 0.015$ to $\pm 0.02$	1.5	$\pm 30$	0 to $-10V$ , $\pm 5V$	Hermetic DIP
DIGITAL-TO-ANALOG CONVERTERS							
Description	Model	Resolution	Linearity Error (% of FSR) max	Tempco (ppm of FSR/ $^{\circ}$ C) max	Output Range	Settling Time (FSR $\pm 1/2$ LSB)	Package
-55 $^{\circ}$ C to +200 $^{\circ}$ C	DAC10HT	12 Bits	$\pm 0.012$ to $\pm 0.048$	$\pm 20$ to $\pm 50$	$\pm 2.5$ , $\pm 5$ , $\pm 10$ , $+5$ , $+10$	200nsec	Hermetic DIP
Very High Speed	DAC60	10 - 12 Bits	$\pm 0.012$ to $\pm 0.048$	$\pm 15$	-5 to $\pm 2.5$ mA	40 - 150nsec	Module
Ultra High Speed	DAC63	12 Bits	$\pm 0.01$	25	0 to $-10$ , $\pm 5$ mA	40nsec	DIP
High Resolution	DAC70	16 Bits, 4 Digits	$\pm 0.003$ to $\pm 0.005$	$\pm 12$ to $\pm 24$	0 to $-2$ mA, $\pm 1$ mA	50 $\mu$ sec	Hermetic DIP
	DAC71	16 Bits, 4 Digits	$\pm 0.003$ to $\pm 0.005$	$\pm 15$ to $\pm 50$	0 to $\pm 10V$ , $-2$ to $\pm 1$ mA	1 - 10 $\mu$ sec	DIP
	DAC72	16 Bits, 4 Digits	$\pm 0.003$ to $\pm 0.005$	$\pm 10$ to $\pm 40$	0 to $\pm 10V$ , $-2$ to $\pm 1$ mA	1 - 10 $\mu$ sec	Hermetic DIP
	DAC73	16 Bits	$\pm 0.00075$ to $\pm 0.0015$	$\pm 24$	$\pm 2.5$ to $\pm 10V$ , $-2$ to $\pm 1$ mA	50 $\mu$ sec	Module
	DAC736	16 Bits	$\pm 0.00075$ to $\pm 0.0015$	$\pm 24$	$\pm 2.5$ to $\pm 10V$ , $-2$ to $\pm 1$ mA	50 $\mu$ sec	Module
High Resolution Monolithic	DAC701/703	16 Bits	$\pm 0.003$	$\pm 18$ to $\pm 28$	0 to $+10V$ , $\pm 10V$	8 $\mu$ sec	Hermetic DIP
	DAC700/702	16 Bits	$\pm 0.003$	$\pm 18$ to $\pm 28$	0 to $-2$ mA, $\pm 1$ mA	300nsec	Hermetic DIP
High Resolution Self-Calibrating	DAC74	16 Bits	$\pm 0.0075$	0 after self-calibration $+15^{\circ}$ C to $+45^{\circ}$ C	0 to $+10V$ , $\pm 10V$	50 $\mu$ sec	Module
Low Cost	DAC80	12 Bits, 3 Digits	$\pm 0.012$ to $\pm 0.025$	$\pm 25$	$\pm 2.5$ to $\pm 10V$ , $-2$ to $\pm 1$ mA	300nsec to 5 $\mu$ sec	DIP
Voltage and Current Output	DAC82	8 Bits	$\pm 0.16$	$\pm 50$	$\pm 2.5$ to $\pm 10V$ , $-1.6$ to $\pm 0.8$ mA	2.5 $\mu$ sec	Hermetic DIP
Low Drift	DAC85	12 Bits, 3 Digits	$\pm 0.012$ to $\pm 0.025$	$\pm 5$ to $\pm 20$	$\pm 2.5$ to $\pm 10V$ , $-2$ to $\pm 1$ mA	300nsec to 5 $\mu$ sec	Hermetic DIP
Monolithic	DAC90	8 Bits	$\pm 0.2$	$\pm 50$	$-2$ to $\pm 1$ mA	200nsec	Hermetic DIP
	DAC800	12 Bits	$\pm 0.012$	$\pm 25$	$-2.5$ to $\pm 10V$ , $-2.35$ to $\pm 1.175$ mA	300nsec to 5 $\mu$ sec	DIP
Very Low Cost Monolithic	DAC800P	12 Bits	$\pm 0.012$	$\pm 25$	$\pm 2.5$ , $\pm 5$ , $\pm 10$ , 0 to $+5$ , 0 to $+10$	3 $\mu$ sec	Plastic DIP
Monolithic	DAC850	12 Bits	$\pm 0.012$	$\pm 17$	$\pm 2.5$ to $\pm 10V$ , $-2.35$ to $\pm 1.175$ mA	300nsec to 5 $\mu$ sec	Hermetic DIP
	DAC851	12 Bits	$\pm 0.012$	$\pm 30$	$\pm 2.5$ to $\pm 10V$ , $-2.35$ to $\pm 1.175$ mA	300nsec to 5 $\mu$ sec	Hermetic DIP

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# DATA CONVERSION [cont]

DIGITAL-AUDIO CONVERTERS						
Description	Model	Resolution (Bits)	Total Harmonic Distortion (max)	Conversion Time (max)	Input Range (V)	Dynamic Range
PCM-Audio A/D Converter <sup>(2)</sup>	PCM75KG	16	0.02% at -15dB	17 $\mu$ sec <sup>(1)</sup>	$\pm 2.5, \pm 5, \pm 10$	90dB
	PCM75JG	14 <sup>(3)</sup>	0.05% at -15dB	15 $\mu$ sec <sup>(1)</sup>	$\pm 2.5, \pm 5, \pm 10$	90dB
Description	Model	Resolution (Bits)	Total Harmonic Distortion (max)	Settling Time <sup>(5)</sup>	Output Range (V)	Dynamic Range
PCM-Audio D/A Converter	PCM50KG	16 <sup>(4)</sup>	0.02% at -15dB	5 $\mu$ sec	$\pm 5, \pm 10$	96dB
Low Cost PCM-Audio D/A Converter	PCM52JG-V/53JG-V	16	0.04% at -20dB	3 $\mu$ sec	$\pm 5, \pm 10$	96dB

NOTES: (1) Can be reduced to 8 $\mu$ sec. (2) With internal 16-bit DAC. (3) Can operate at 16 bits. (4) Can operate at 14 bits. (5) 20V step.

VOLTAGE-TO-FREQUENCY CONVERTERS						
Description	Model	V <sub>IN</sub> Range (V)	Four Range (kHz)	Linearity (% of FSR) max	Tempco (ppm of FSR/°C) max	Package
High Linearity Monolithic	VFC320	0 to +10	0 to 1000	$\pm 0.005\%$ at 10kHz $\pm 0.03\%$ at 100kHz $\pm 0.1\%$ typ at 1MHz	$\pm 20$ to $\pm 50$	DIP & TO-100
High Linearity Monolithic Totem Pole Output	VFC62	0 to +10	0 to 1000	$\pm 0.005\%$ at 10kHz $\pm 0.03\%$ at 100kHz $\pm 0.1\%$ typ at 1MHz	$\pm 20$ to $\pm 50$	DIP & TO-100
Low Cost, Monolithic	VFC32	0 to +10	0 to 500	$\pm 0.01$ at 10kHz $\pm 0.05$ at 100kHz	$\pm 75$ to $\pm 150$	DIP & TO-100
Low Cost, Complete Hybrid	VFC42	0 to +10	0 to 10	$\pm 0.01$	$\pm 100$	DIP
	VFC52	0 to +10	0 to 100	$\pm 0.05$	$\pm 150$	DIP
Low Drift, Complete	VFC12	0 to +10	0 to 10	$\pm 0.01$	50	Module
	VFC15	0 to +20	0 to 20	$\pm 0.01$	50	Module

MICROPROCESSOR-INTERFACED ANALOG INPUT SYSTEMS						
Compatible With	Model	Channels	Resolution (Bits)	Accuracy (% of FSR) max	Accuracy Drift (ppm/°C) max	Package
8080, SC/MP Compatible	MP20	8 Differential, 16 Single-ended	8	$\pm 0.8$ at high gain, $\pm 0.4$ at low gain	$\pm 40$	DIP
6800, 6502 Compatible	MP21	8 Differential, 16 Single-ended	8	$\pm 0.8$ at high gain, $\pm 0.4$ at low gain	$\pm 40$	DIP
Universal	MP22	8 Differential, 16 Single-ended	12	$\pm 0.4$ at high gain, $\pm 0.1$ at low gain	$\pm 25$	DIP
	MP32	8 Differential, 16 Single-ended	12	$\pm 0.4$ at high gain, $\pm 0.05$ at low gain	$\pm 25$	DIP

MICROPROCESSOR-INTERFACED ANALOG OUTPUT SYSTEMS						
Compatible With	Model	Channels	Resolution (Bits)	Accuracy (% of FSR)	Accuracy Drift (ppm/°C)	Package
8080, SC/MP Compatible	MP10	2	8	$\pm 0.4$	$\pm 80$	DIP
6800, 6502 Compatible	MP11	2	8	$\pm 0.4$	$\pm 80$	DIP

DATA ACQUISITION SYSTEMS						
Description	Model	Channels	Resolution (Bits)	Accuracy (% of FSR)	Throughput Rate (kHz)	Package
Modular, Low Level	SDM853	8 Differential, 16 Single-ended	12	$\pm 0.025$	30	Module
Hybrid, $\pm 10V$ Input	SDM854	8 Differential, 16 Single-ended	12	$\pm 0.024$ to $\pm 0.048$	25 to 33	DIP
Hybrid, $\pm 5V$ Input	SDM856	8 Differential, 16 Single-ended	12	$\pm 0.024$ to $\pm 0.048$	25 to 33	DIP
Hybrid, Low Level	SDM857	8 Differential, 16 Single-ended	12	$\pm 0.024$ to $\pm 0.048$	18 to 22	DIP
Modular, Low Level	SDM858	8 Differential, 16 Single-ended	12	$\pm 0.025$ (at G = 100)	8	Module

SAMPLE/HOLD CIRCUITS							
Description	Model	Gain Offset Error (%) (mV)	Charge Offset (mV)	Droop Rate (mV/msec)	Tempco (ppm of 20V/°C)	Acquisition Time ( $\mu$ sec)	Package
Low Cost, Complete	SHC80	$\pm 0.01, \pm 2$ max	$\pm 2$ max	0.5max	$\pm 3$	10 max	DIP
High Speed, Complete	SHC85	$\pm 0.01, \pm 2$ max	$\pm 2$ max	0.5 max	$\pm 3$	4.5 max	Hermetic DIP
Low Cost, Monolithic	SHC298	$\pm 0.01$ max, $\pm 7$ max	$\pm 25$ max	0.125 max	$\pm 4$	10 max	TO-99
Very High Speed	SHM60	$\pm 0.01$ max, $\pm 1.5$	$\pm 1.5$	0.005/0.001	$\pm 2$	1 max	Module

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## DATA CONVERSION [cont]

MULTIPLEXERS							
Type	Model	Channels	Input Range (V)	On Resistance (k $\mu$ ) max	Crosstalk (% of Off Channel Signal)	Settling Time (to 0.01%) ( $\mu$ sec)	Package
CMOS	MPC8S	8 Single-ended	$\pm 15$	1.8	0.005	5	Hermetic DIP
	MPC4D	4 Differential	$\pm 15$	1.8	0.005	5	Hermetic DIP
	MPC16S	16 Single-ended	$\pm 15$	1.8	0.005	7	Hermetic DIP
	MPC8D	8 Differential	$\pm 15$	1.8	0.005	7	Hermetic DIP
CMOS (-55 to +125°C)	MPC800	16 Single-ended, 8 Differential	$\pm 15$	0.75	0.0005	0.8	Hermetic DIP
	MPC801	8 Single-ended, 4 Differential	$\pm 15$	0.75	0.0005	0.8	Hermetic DIP

## OPERATIONAL AMPLIFIERS

LOW DRIFT									
Description	Model	Rated Output		Offset Voltage		Bias Current at 25°C pA	Frequency Response		Package
		$\pm V$	$\pm mA$	at 25°C mV	Temp. Drift $\mu V/^\circ C$		Unity Gain MHz	Slew Rate V/ $\mu$ sec	
Low Power	OPA21	13.5	1.3	0.1 - 0.5	1-5	50 - 100nA	0.6	0.25	TO-99, DIP
Low Noise	OPA27	12	11.5	0.025 - 0.1	0.6 - 1.8	40 - 80nA	5	1.7	TO-99, DIP
Low Noise (AV = 5 min)	OPA37	12	11.5	0.025 - 0.1	0.6 - 1.8	40 - 80nA	40	11	TO-99, DIP
Switchable-Input (SWOP-Amp)	OPA201	$\pm 10$	$\pm 5$	0.1 - 0.5	1 - 5	25 - 50nA	0.5	0.1 min	DIP
Chopper-Stabilized	3291	10	5	0.02 - 0.10	0.10 - 1	$\pm 50$ to $\pm 100$	3	6	Module
	3354	10	5	0.03 - 0.10	0.10 - 1	$\pm 20$ to $\pm 50$	3	6	Module
	3271	50 - 110	20	0.05	1	$\pm 80$	1	20	Module
Bipolar	3500	10	10	0.5 - 5	1 - 20	$\pm 15$ to $\pm 50nA$	1.5	0.6 - 1	TO-99
	3501	10	5	2 - 5	5 - 20	$\pm 3$ to $\pm 15nA$	0.5	0.1	TO-99
	3510	10	10	0.06 - 0.15	0.5 - 2	$\pm 15$ to $\pm 35nA$	0.4	0.5	TO-99
FET	3521	10	10	0.25 - 0.50	1 - 10	-10 to -20	1.5	0.6	TO-99
	3527	10	10	0.25 - 0.50	2 - 10	-2 to -5	1	0.6	TO-99
LOW BIAS CURRENT									
-55°C to +200°C	OPA11HT	10	15	5	5	$\pm 25nA$	12	7	TO-99
Low Bias, Low Drift, Enhanced Bi FET	OPA100	10	5	0.25 - 1	5 - 15	$\pm 1$ to $\pm 3$	1	2	TO-99, DIP
Low Noise	OPA101/102	12	12	0.25 - 0.5	5 - 10	-10 to -15	10	5 - 10	TO-99
Low Bias	OPA103	10	5	0.25 - 0.5	2 - 25	-1 to -2	1	0.9	TO-99
Ultra-Low Bias	OPA104	10	5	0.5 - 1.0	10 - 25	-0.075 to -0.3	1	1.6	TO-99
FET Ultra-Low Bias	3528	10	5	0.25 - 0.50	5 - 15	$\pm 0.075$ to -0.30	0.7	0.3	TO-99
Low Bias	3527	10	10	0.25 - 0.50	5 - 15	-2 to -5	1	0.6	TO-99
Low Bias	3522	10	10	0.5 - 1	10 - 50	-1 to -10	1	0.6	TO-99
Low Drift	3521	10	10	0.25 - 0.50	1 - 10	-10 to -20	1.5	0.6	TO-99
Low Cost	3542	10	10	20	50	-25	1	0.5	TO-99
Varactor	3430	10	5	Adj. to 0	10 - 30	$\pm 0.01$	2kHz	0.4V/msec	Module
	3431	10	5	Adj. to 0	10 - 30	$\pm 0.01$	2kHz	0.4V/msec	Module
HIGH VOLTAGE									
FET	3584	65 - 145	15	3	25	-20	7	150	TO-3
	3583	40 - 140	75	3	25	-20	5	30	TO-3
	3580	10 - 145	15 - 60	3 - 10	25 - 30	-20 to -50	5	15 - 20	TO-3
HIGH CURRENT									
High Power	OPA501	32	10A	5 - 10	40 - 65	20 - 40nA	1	1.5	TO-3
FET	3573	20	2A	10	65	40nA	1	1.5	TO-3
	3572	30	2A	2	40	-100	0.5	3	TO-3
	3571	30	1A	2	40	-100	0.5	3	TO-3
Buffer	3553	10	200	50	300	-200	32	2000	TO-3
	3329	10	100	50	-	Bipolar	1	63	DIP

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WIDEBAND										
Description	Model	Rated Output		Offset Voltage		Frequency Response		ts ±0.1% nsec	Compen- sation	Package
		±V	±mA	at 25°C mV	Temp. Drift µV/°C	A - BW MHz	Slew Rate V/µsec			
-55° to +200°C	OPA12HT	10	10	10	30	20, A = 10	80	200	int.	TO-99
Fast Settling	OPA605	10	30	0.5 - 1.0	5 - 25	200, A = 1000	300	500	ext.	DIP
Differential	3554	10	100	1 - 2	15 - 50	1000, A = 1000	1000	120	ext.	TO-3
	3551	10	10	1	50	50, A = 10	250	400	ext.	TO-99
	3550	10	10	1	50	10 - 20, A = 1	65 - 100	400	int.	TO-99
	3508	10	10	5	30	100, A = 100	20	—	ext.	TO-99
	3507	10	10	10	30	20, A = 10	80	200	ext.	TO-99
Buffer	3553	10	200	50	300	32 <sup>(1)</sup>	2000	—	—	TO-3

(1) Full power bandwidth

**INSTRUMENTATION AMPLIFIERS**

LOW DRIFT								
Description	Model	Gain		Rated Output V/mA	Input Parameters		Dynamic Response G = 100 ±3dB, BW	Package
		Range (V/V)	Non-Linearity G = 100, max		CMR, DC to 60Hz, G = 10 1kΩ Unbal.	Offset Voltage vs. Temp. max (µV/°C)		
Monolithic, Low Cost Very High Accuracy	INA101	1 - 1000	±0.003%	±10/±5	96dB, min	±0.25 ±10/G	25kHz	TO-100, DIP
Low Drift, Low Cost	3636	5 - 1000	±0.04%	±10/±5	80dB, min.	±(1 + 5/G)	14kHz	DIP
	3629	5 - 1000	±0.004%	±10/±10	90dB, min.	±(0.75 + 5/G)	30kHz	DIP
General Purpose, Low Cost	3660	1 - 1000	±0.03%	±10/±10	90dB, min.	±(2 + 500/G)	72kHz	TO-100
	3662	1 - 1000	±0.05%	±10/±10	84dB, min.	±(2 + 400/G)	74kHz	DIP
Buffer, Unity Gain, Differential	3627	1 (fixed)	±0.001% G = 1	±10/±5	100dB, min.	20	800kHz G = 1	TO-99
Very High Accuracy	3630	1 - 1000	±0.003%	±10/±5	96dB, min.	±(0.25 + 10/G)	25kHz	DIP

**PROGRAMMABLE GAIN**

Digitally Controlled (4-Bit TTL Input)	3606 <sup>(1)</sup>	1 - 1024 (11 Gains)	0.004%	±10/±5	100dB, min.	±(1 + 20/G)	40kHz	DIP
Digitally Controlled Programmable Gain Multiplexed Input	PGA100	1 - 128 (8 Gains)	±0.005%	±10/±2	—	±6	50kHz	DIP

**4 to 20mA TRANSMITTER**

Description	Model	Input V	Non-Linearity, max	Output	Input CMRR	Input V <sub>os</sub> vs Temp	Input V <sub>os</sub>	Package
Precision Self-Powered Two-Wire Transmitter	XTR100	0 - 1	0.01%	4 - 20mA	90dB min.	±0.5µV/°C	±25µV	DIP

(1) 3606 offers software gain control with "G" set by 4-bit TTL inputs.

**ISOLATION AMPLIFIERS**

TRANSFORMER COUPLED AMPLIFIERS															
Description	Model	Input Voltage		Input Impedance		Isolation Voltage		Isolation Mode		Leakage Current at 240VAC (µA)	Gain Nonlinearity		±3dB Freq. (kHz)	External Isolation Power Required	Package
		Common Mode (V)	Differ-ential (V)	Common Mode (Ω)	Differ-ential (Ω)	Contin-uous (V) Peak	Pulse/ Test (V) Peak	Rejection, min.							
								DC (dB)	60Hz (dB)		Max (%)	Typ (%)			
Low Drift, Bipolar	3450	±10	±15	5 × 10 <sup>9</sup>	10 <sup>7</sup>	±500	±2000	160	120	2.5	±0.005	±0.0015	1.5	No	Module
Low Bias, FET	3451	±10	±15	10 <sup>11</sup>	10 <sup>11</sup>	±500	±2000	160	120	2.5	±0.025	±0.005	2.5	No	Module
	3452	±10	±15	10 <sup>11</sup>	10 <sup>11</sup>	±2000	±5000	160	120	2.5	±0.025	±0.005	2.5	No <sup>(1)</sup>	Module
	3455	±10	±15	10 <sup>11</sup>	10 <sup>11</sup>	±2000	±5000	160	120	2	±0.025	±0.005	2.5	No <sup>(1)</sup>	Module
True 3-wire Inst. Amp.	3456	±10	±20	5 × 10 <sup>9</sup>	10 <sup>7</sup>	±2000	±5000	160	130	2.5	±0.02	±0.01	2.5	No <sup>(1)</sup>	Module
IC with Power Supply	3656	±8	±8	10 <sup>8</sup>	10 <sup>8</sup>	±3500	±8000	160	125	1	±0.05 to ±0.15	±0.03	30	No <sup>(1)</sup>	DIP

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## ISOLATION AMPLIFIERS [cont]

OPTICALLY COUPLED AMPLIFIERS															
Description	Model	Input Voltage		Input Impedance		Isolation Voltage		Isolation Mode Rejection, min.		Leakage Current at 240VAC ( $\mu$ A)	Gain Nonlinearity		$\pm$ 3dB Freq. (kHz)	External Isolation Power Required	Package
		Common Mode (V)	Differential (V)	Common Mode ( $\Omega$ )	Differential ( $\Omega$ )	Continuous (V) Peak	Pulse/Test (V) Peak	DC (dB)	60Hz (dB)		Max (%)	Typ (%)			
Miniature Low Cost	ISO100					$\pm$ 750	$\pm$ 2500	5pA/V	400pA/V	0.3	0.07	0.02	60	Yes <sup>(4)</sup>	DIP
Balanced Current Input	3650	$\pm$ 10	$\pm$ 15	10 <sup>9</sup>	25 <sup>(3)</sup>	$\pm$ 2000	$\pm$ 5000	140	120	0.35	$\pm$ 0.05 to $\pm$ 0.2	$\pm$ 0.02 to $\pm$ 0.05	15	Yes <sup>(4)</sup>	DIP
Balanced FET Input	3652	$\pm$ 10	<sup>(5)</sup>	10 <sup>11</sup>	10 <sup>11</sup>	$\pm$ 2000	$\pm$ 5000	140	120	0.35	$\pm$ 0.1 to $\pm$ 0.2	$\pm$ 0.05	15	Yes <sup>(4)</sup>	DIP

NOTES: (1) Isolated power available to power external circuitry. (2) Isolation voltage tested at both  $\pm$ 5000Vp pulse test and 2500V, rms, 60Hz; leakage current is 2 $\mu$ A max, at 240V, rms, 60Hz. (3) 3650 is a current input device: to reduce input errors, very low input impedance is desirable. (4) Models 722 or 724 isolated DC/DC converters may be used to provide isolated power. (5) Protected up to  $\pm$ 6000V.

## VOLTAGE REFERENCE

Description	Model	Output (V)	Minimum Output (mA)	Maximum Drift (ppm/ $^{\circ}$ C)	Power Supply		Package
					(V)	(mA)	
Precision	REF101	+10.000 $\pm$ 0.005	10	1 - 6	+13.5 to +35	4.5	TO-99

## ANALOG CIRCUIT FUNCTIONS

MULTIPLIER/DIVIDERS			
Model	Accuracy at 25 $^{\circ}$ C, max, No Trimming Required	Specification Temp. Range ( $^{\circ}$ C)	Package
4214	0.50 - 1%	-25 to +85, -55 to +125	DIP
4203/05	1 - 2%	0 to +70, -55 to +125	TO-100
4204	0.25 - 0.50%	-25 to +85, -55 to +125	DIP
4206	0.25% - 0.50%	0 to +70	DIP
4213	0.50 - 1%	-25 to +85, -55 to +125	TO-100

SPECIAL FUNCTIONS			
Description	Model	Comments	Package
Logarithmic Amplifier	4127	0.50 - 1% accuracy, 2 current or 2 voltage input signals	DIP
Log Ratio Amplifier	LOG100	0.1% - 0.37% accuracy, 2 current input log ratio function	DIP
Divider	DIV100	$E_o = 10 N/D$ , $\pm$ 0.25 - 1% untrimmed accuracy	DIP
Multifunction Converters	4301	$E_o = V_i(V_z/V_x)^m$ , hermetically sealed metal package, compatible with 4302	DIP
Multiply, divide, square root, etc.	4302	$E_o = V_i(V_z/V_x)^m$ , plastic package	
Peak Detector	4085	Provide read-out of DC voltage equal to peak value of complex wave form	DIP
RMS-to-DC Converters	4340 4341	Laser-trimmed, no external trimming, metal package, pin compatible with 4341 Externally trimmed, low cost in plastic package	DIP DIP
Oscillators	4023/25 4423	Fixed-frequency (10Hz to 20kHz) Very-low cost, RC programmable, 0.002Hz to 20kHz	Module DIP
Comparators	4115/04 4082/03	Provides window or dual limit for comparison Drives lamps, relays or logic loads	Module Module
Fixed-Frequency Active Filters	ATF76 Series	Pre-tuned, multiple pole bandpass, low-pass, and notch filters	Module
Universal Active Filters	UAF11/21/31/41	Variable type, 4-resistor, programmable, 2-pole filters; Q and $F_o$ can be programmed; 3 outputs provide low-pass, high-pass and bandpass transfer functions; complex filter response by cascading.	DIP

FOR ADDITIONAL DETAILS REFER TO THE BURR-BROWN PRODUCT DATA BOOK OR CONTACT YOUR NEAREST BURR-BROWN SALES OFFICE.



# MILITARY PRODUCTS

Burr-Brown's Military products meet the demand for high quality products for high reliability applications. They feature exceptional electrical performance from  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ , Hi-Rel manufacture, MIL-STD-883 class B screening, and reasonable prices.

They are designed to conservative and stringent MIL-M-38510 requirements, and are produced on a separate manufacturing line, which exceeds Burr-Brown's standard lines' quality. MIL-M-38510 and MIL-STD-883 processes and controls are used throughout.

INTERFACE  
Burr-Brown

ANALOG-TO-DIGITAL CONVERTERS										
Model	Resolution Bits	Linearity $\pm\text{LSB}$ , max	Conversion Time $\mu\text{sec}$ , max	Gain Drift $\pm\text{ppm}/^{\circ}\text{C}$ , max	Input Range V		Package			
ADC87/MIL, /883B ADC87U/883B, U	12	1/2	8	15	$\pm 2.5, \pm 5, \pm 10, 0$ to $+5, 0$ to $+10$		32-pin DIP			
DIGITAL-TO-ANALOG CONVERTERS										
Model	Resolution Bits	Linearity $\pm\text{LSB}$ , max	Monotonicity	Gain Drift $\pm\text{ppm}/^{\circ}\text{C}$ , max	Settling Time max	Output Ranges	Package			
DAC87-CBI-V/MIL, -CBI-V/B, -CBI-V	12	1/2	$-55^{\circ}\text{C}/+125^{\circ}\text{C}$	20	7 $\mu\text{sec}$	$\pm 2.5, \pm 5, \pm 10,$ $+5, +10\text{V}$	24-pin DIP			
DAC87U-CBI-V/B, -CBI-V	12	1/2	$-25^{\circ}\text{C}/+85^{\circ}\text{C}$	20	7 $\mu\text{sec}$		24-pin DIP			
DAC87-CBI-I/B	12	1/2	$-55^{\circ}\text{C}/+125^{\circ}\text{C}$	20	400nsec	$0$ to $-2\text{mA}$ $\pm 1\text{mA}$	24-pin DIP			
DAC87-CBI-I	12	1/2	$-55^{\circ}\text{C}/+125^{\circ}\text{C}$	20	400nsec		24-pin DIP			
DAC87U-CBI-I/B	12	1/2	$-25^{\circ}\text{C}/+85^{\circ}\text{C}$	20	400nsec		24-pin DIP			
DAC87U-CBI-I	12	1/2	$-25^{\circ}\text{C}/+85^{\circ}\text{C}$	20	400nsec		24-pin DIP			
DAC870V/MIL, /883B, /(none)	12	1/2	$-55^{\circ}\text{C}/+125^{\circ}\text{C}$	25	7 $\mu\text{sec}$	$\pm 2.5, \pm 5, \pm 10,$ $+5, +10\text{V}$	24-pin DIP			
DAC870U/883B, (none)	12	1/2	$-25^{\circ}\text{C}/+85^{\circ}\text{C}$	20	7 $\mu\text{sec}$		24-pin DIP			
DAC870VL/MIL, 883B, /(none)	12	1/2	$-55^{\circ}\text{C}/+125^{\circ}\text{C}$	25	7 $\mu\text{sec}$		28-terminal LCC			
DAC870UL/883B, /(none)	12	1/2	$-25^{\circ}\text{C}/+85^{\circ}\text{C}$	20	7 $\mu\text{sec}$		28-terminal LCC			
8300201XC*	12	1/2	$-55^{\circ}\text{C}/+125^{\circ}\text{C}$	20	7 $\mu\text{sec}$	$\pm 2.5, \pm 5, \pm 10,$ $+5, +10\text{V}$	24-pin DIP			
* DESC Drawing 83002 to DAC87-CBI-V/MIL.										
MULTIPLIERS										
Model	Accuracy at $25^{\circ}\text{C}$ $\pm\%$ , max	Accuracy at $125^{\circ}\text{C}$ $\pm\%$ , max	Feedthrough $\pm\text{mV}$ , max	Output Offset $\pm\text{mV}$ , max	Output V, mA, min	Package				
4213WM/883B, WM	1/2	4	50	25	$\pm 10, \pm 5$	TO-100				
4213VM/MIL, VM/883B, VM	1	4	100	30	$\pm 10, \pm 5$	TO-100				
4213UM/883B, UM	1	2 <sup>(1)</sup>	100	50	$\pm 10, \pm 5$	TO-100				
VOLTAGE-TO-FREQUENCY CONVERTERS										
Model	V <sub>IN</sub> Range V	F <sub>OUT</sub> Range kHz, max	Linearity % FSR, max	Full Scale Drift ppm FSR/ $^{\circ}\text{C}$ , max	Package					
VFC32WM/883B, WM	$\pm 10$	200	$\pm 0.006$ at 10kHz	$\pm 100$ at 10kHz	TO-100					
VFC32VM/MIL, VM/883B, VM	$\pm 10$	200	$\pm 0.01$ at 10kHz	$-400, +150$ at 200kHz	TO-100					
VFC32UM/883B, UM	$\pm 10$	200	$\pm 0.01$ at 10kHz	$\pm 150$ at 10kHz	TO-100					
OPERATIONAL AMPLIFIERS										
Description	Model	Offset Voltage		Bias Current nA, max	Bandwidth Unity Gain MHz, min	Slew Rate V/ $\mu\text{s}$ , min	t <sub>s</sub> $\pm 0.01\%$ ns	Compensation	V, mA, min	Package
		at $25^{\circ}\text{C}$ $\pm\text{mV}$ , max	drift $\pm\mu\text{V}/^{\circ}\text{C}$ max							
Wideband	OPA600VM/MIL, VM/883B, VM	2	20	-100pA	5000 <sup>(2)</sup> , A = 1000	400	125	external	$\pm 10, \pm 200$	16-pin DIP
	OPA600UM/883B, UM	5	60	-100pA	5000 <sup>(2)</sup> , A = 1000	400	150	external	$\pm 10, \pm 200$	16-pin DIP
General Purpose Bipolar	3500R/MIL, /883B	5	20	$\pm 30$	1	0.6	—	internal	$\pm 10, \pm 10$	TO-99
	3500U/883B	5	20 <sup>(3)</sup>	$\pm 30$	1	0.6	—	internal	$\pm 10, \pm 10$	TO-99
Precision Bipolar	3510VM/MIL, /883B	0.12	2	$\pm 25$	0.25	0.5	—	internal	$\pm 10, \pm 10$	TO-99
Low Drift, Low Bias	OPA105WM/MIL, WM/883B, WM	.250	2	-1pA	1	0.9	—	internal	$\pm 10, \pm 10$	TO-99
	OPA105VM/MIL, VM/883B, VM	.250	5	-1pA	1	0.9	—	internal	$\pm 10, \pm 10$	TO-99
	OPA105UM/883B, UM	.250	15 <sup>(3)</sup>	-1pA	1	0.9	—	internal	$\pm 10, \pm 10$	TO-99
Ultra Low Bias	OPA106WM/MIL, WM/883B, WM	.250	5	-100fA	1	1.2	—	internal	$\pm 10, \pm 5$	TO-99
	OPA106VM/MIL, VM/883B, VM	.250	10	-150fA	1	1.2	—	internal	$\pm 10, \pm 5$	TO-99
	OPA106UM/883B, UM	.250	20 <sup>(3)</sup>	-300fA	1	1.2	—	internal	$\pm 10, \pm 5$	TO-99

NOTES: (1) At  $+85^{\circ}\text{C}$ . (2) Gain-bandwidth product. (3)  $-25^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ .

FOR ADDITIONAL DETAILS REFER TO THE BURR-BROWN PRODUCT DATA BOOK OR CONTACT YOUR NEAREST BURR-BROWN SALES OFFICE.

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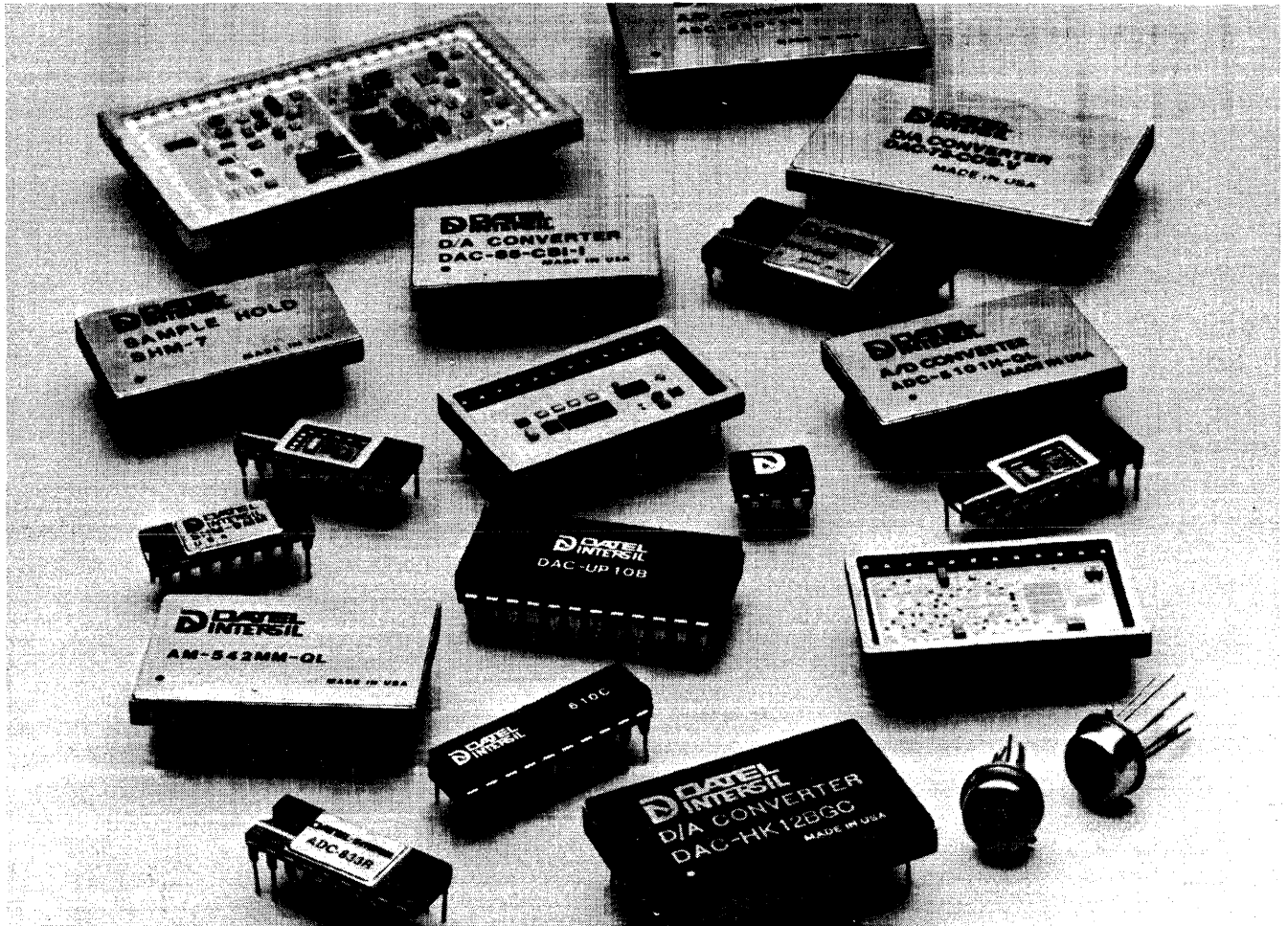
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DateI offers one of the broadest lines of data-conversion products in the industry. Included are A/D and D/A converters, sample-and-holds, operational and instrumentation amplifiers, and data-acquisition systems — all packaged using the latest in thin-film, hybrid-microelectronic-circuit manufacturing technologies.

DateI also offers fast delivery. On standard products, delivery typically runs from stock to four weeks and from six to eight weeks for full military-grade products with Class 883B screening.

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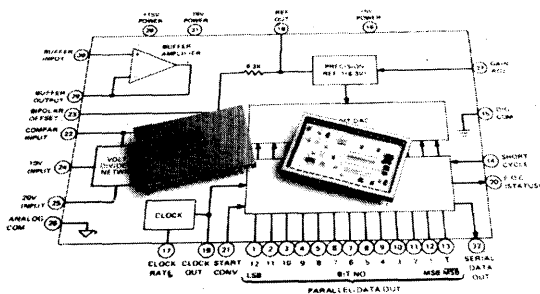


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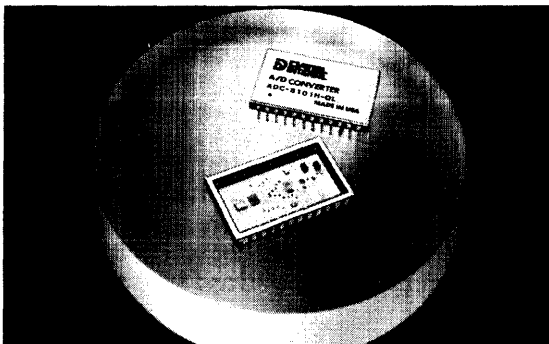
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# New Products from Datel



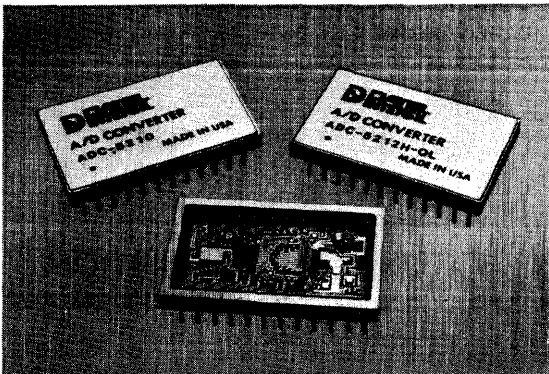
## ADC-810, ADC-811 Ultra-fast, 12-bit Hybrid Analog to Digital Converters

The ADC-810 and ADC-811 are ultra-fast, hybrid, successive approximation, 12-bit analog to digital converters. The ADC-810 achieves a maximum conversion time of 2  $\mu$ sec. Conversion time for the ADC-811 is 4  $\mu$ sec maximum, which is the only difference between the two. Both models are pin-compatible with industry-standard ADC-85/87 converters, offering increased speed, high accuracy and reliability. Models are available subjected to 100% screening to MIL-STD-883B.



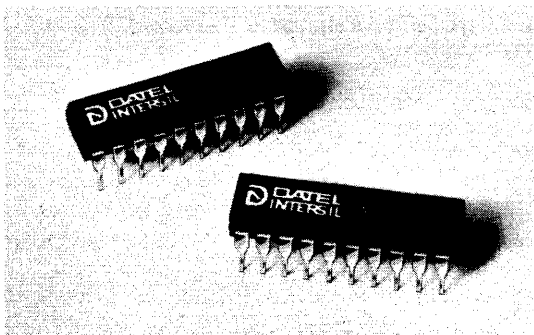
## ADC-5101 high speed, 8-bit A/D offers high temperature operation at low cost

The ADC-5101 is a high speed, adjustment-free, 8-bit A/D converter. Pin compatible with standard ADC-5101 converters, these devices offer high accuracy and high speed over the full military operating temperature range of  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ . Designed for operation without external adjustment circuits, the ADC-5101 accomplishes an 8-bit conversion in only 900 ns maximum. Models are available subjected to 100% screening to MIL-STD-883 Class B.



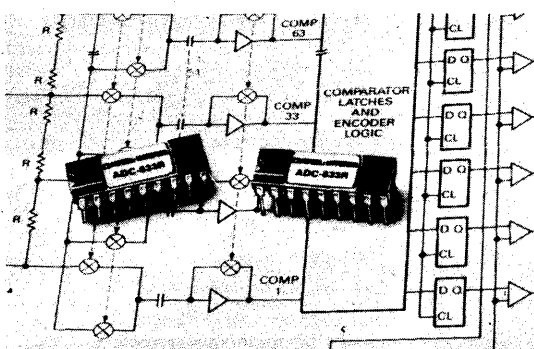
## ADC-5210 series adjustment-free 12-bit A/D features high accuracy over military temperature range

The ADC-5210 series are high performance, 12-bit successive approximation A/D converters. Completely pin and function compatible with standard ADC-5210 devices, these models offer significantly improved high-temperature operation at lower cost. Full scale absolute accuracy error is a maximum of  $\pm 0.05\%$  FSR at  $+25^{\circ}\text{C}$  and only  $\pm 0.2\%$  FSR over the full military operating temperature range of  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ; an improvement of 10 LSBs over the error specified on competing devices. MIL-STD-883 screening is available.



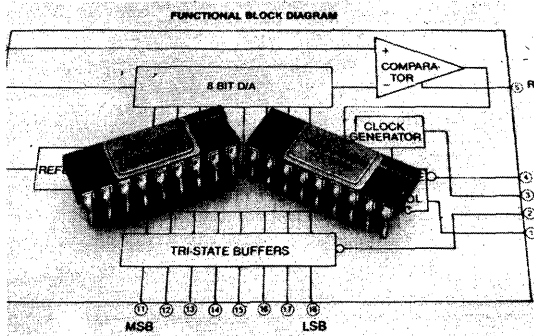
**ADC-830 microprocessor compatible 8-bit A/D converter**

DATEL's ADC-830 is a low-cost, monolithic 8-bit A/D converter designed to operate directly with the 8080A control bus via three-state outputs. The device appears as a memory location or I/O port to the microprocessor and thus does not require interfacing logic. Using the successive approximation technique and a modified potentiometric resistor ladder, the ADC-830 achieves an 8-bit conversion in 100  $\mu$ s with a maximum total adjusted error of only  $\pm 1/2$  LSB. Its combination of low cost, small size, and interface versatility make the ADC-830 an ideal choice for many process control and instrumentation applications.



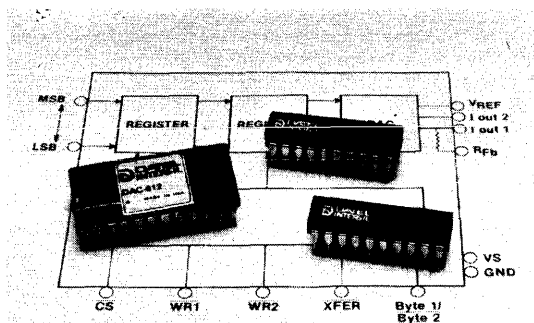
**ADC-833 6-bit A/D features video speed at low power**

The ADC-833 is a low-power, video-speed, 6-bit flash A/D manufactured with CMOS/SOS technology. The device is capable of digitizing an analog input signal at conversion rates up to 15MHz while its power consumption is only 200mW. The analog input voltage range is +2.5V to +10V, and typical differential linearity error is only  $\pm 1/2$  LSB. Outputs are buffered three-state and include an overflow output which allows the user to cascade two units to achieve 7-bit resolution.



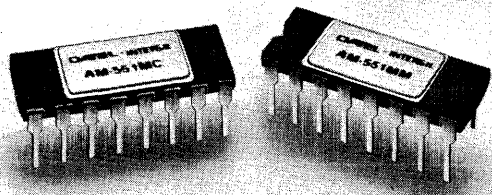
**ADC-847 Microprocessor Compatible, 8-Bit A/D Converter**

The ADC-847 is a low cost, monolithic, 8-bit A/D converter designed to interface directly to a microprocessor via three-state outputs. Utilizing the successive approximation technique, the ADC-847 completes an 8-bit conversion in 9  $\mu$ S with a maximum linearity error as low as  $\pm 1/4$  LSB. The ADC-847 includes, an internal clock generator, reference circuit, successive approximation register, comparator, three-state buffers, 8 bit D/A converter and interface and control logic, making it an ideal choice for many process control and instrumentation applications. The ADC-847 is available for operation over the commercial, 0°C to +70°C and military, -55°C to +125°C temperature ranges and is packaged in either an 18 pin plastic or ceramic DIP.



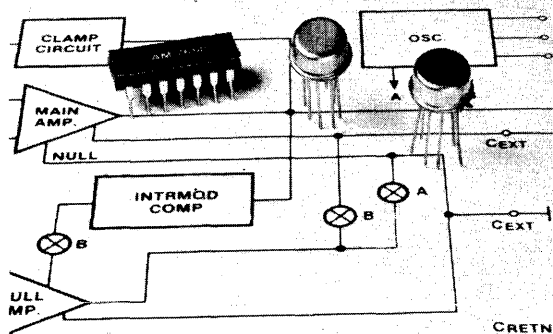
**DAC-608/610/612 Low Cost, Double-Buffered Microprocessor Compatible D/A Converters**

The DAC-608, DAC-610 and DAC-612 are low cost monolithic 8, 10 and 12 bit multiplying D/A converters, designed to interface directly with most popular microprocessors. Double-Buffered inputs allow the converters to output an analog voltage corresponding to one digital word while holding the next. The converters consist of an input register, a D/A register, control logic and D/A converter. Settling time is 1  $\mu$ s for the DAC-608 and DAC-612 and 500 nS for the DAC-610. The reference input of these converters is selectable over a range of  $\pm 10$ V and may also be used as the analog input for four quadrant multiplication applications.



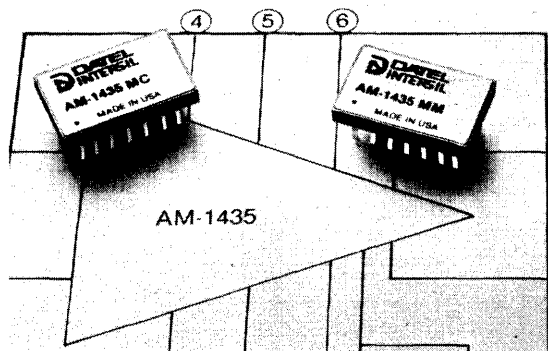
**AM-551 Low Cost, Hybrid Programmable Instrumentation Amplifier**

DATEL's AM-551 is a low cost, high performance, programmable gain instrumentation amplifier manufactured with hybrid thin-film technology. Gain range is 1 to 1000 and is set externally by a single resistor and simple pin-strapping. Settling time is 2  $\mu$ sec for a 20V output step to 0.01% accuracy, with a slew rate of 23V/ $\mu$ sec and a small signal bandwidth of 400 kHz. Applications include remote amplification of low-level signals produced by thermocouples, strain gages and RTD's, high performance data acquisition systems and remote instrumentation systems.



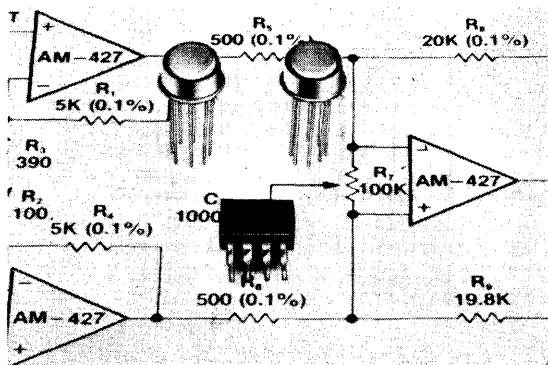
**AM-7650 new, low cost chopper stabilized operational amplifier features ultra-low offset voltage and drift**

DATEL's AM-7650 is a low cost, high performance chopper stabilized amplifier featuring exceptionally low offset voltage and input bias specifications combined with excellent bandwidth and speed characteristics. Input offset voltage is typically  $\pm 0.7 \mu$ V with an input offset voltage drift as low as  $0.01 \mu$ V/ $^{\circ}$ C. The AM-7650 achieves its low offset by comparing the inverting and non-inverting input voltages in a nulling amplifier, which is nulled by alternate clock phases from the internal clock oscillator. The 14 pin DIP version also includes a provision for the use of an external clock if an application requires. These devices are internally compensated for unity-gain operation.



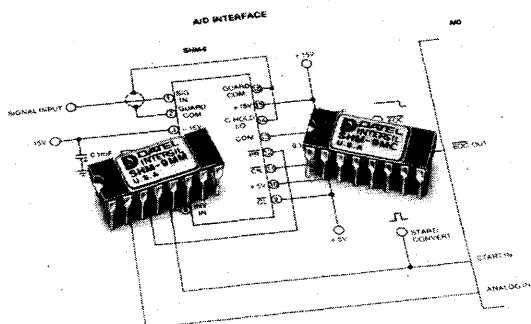
**AM-1435 Ultra High Speed, Wideband, Hybrid Operational Amplifier**

DATEL's AM-1435 is an ultra-fast settling, wide-band operational amplifier. The AM-1435 achieves a settling time of only 70 nsec for a 10V step to 0.01% accuracy. High speed performance is optimized with high open loop gain, flat frequency response beyond 10 kHz and a roll-off of 6 dB/octave to beyond 100 MHz. Gain bandwidth product is typically 1GHz and slew rate is 300V/ $\mu$ sec. All models are packaged in hermetically sealed ceramic cases, with full screening to MIL-STD-883 level B available.



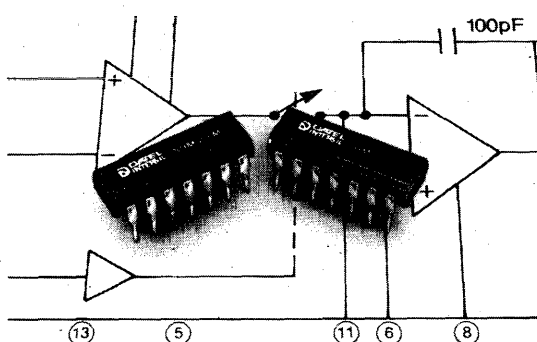
**AM-427 Ultra-Low Noise Monolithic Operational Amplifier**

The AM-427 is a low cost, ultra-low noise operational amplifier which is designed for instrumentation grade signal conditioning where low noise, low offset voltage and low offset voltage drift are required. The AM-427 combines exceptional DC performance with ultra-low noise operation. Maximum input noise voltage density and input noise current density at 10 Hz are 5.5 nV/ $\sqrt$ Hz and 0.4 pA/ $\sqrt$ Hz respectively. The AM-427 is available for operation over the industrial,  $-25^{\circ}$ C to  $+85^{\circ}$ C and military,  $-55^{\circ}$ C to  $+125^{\circ}$ C temperature ranges and is packaged in either an 8 pin TO-99 or an 8 pin ceramic DIP.



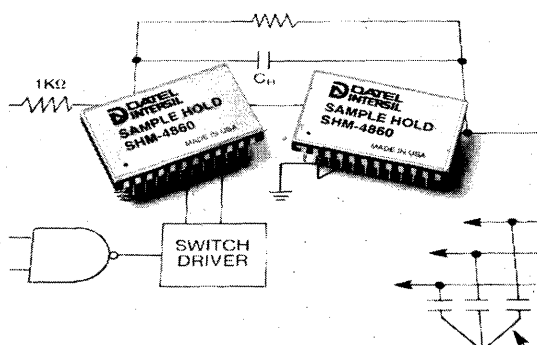
**Low cost functionally complete hybrid sample-hold features .01% accuracy SHM-9**

The SHM-9 is a complete, self-contained sample-hold amplifier that combines high performance versatility with low cost. The SHM-9 includes a bipolar input amplifier, a low-leakage electronic switch, a FET output amplifier, a precision 1000 pF hold capacitor and logic control circuitry. The internal control circuitry allows the SHM-9 to be interfaced with virtually any A/D converter using the converter's Start/Convert and E.O.C. (status) signals. Active laser trimming of highly stable thin-film resistor networks minimizes offset and sample-to-hold offset errors, eliminating the need for external adjustment circuits.



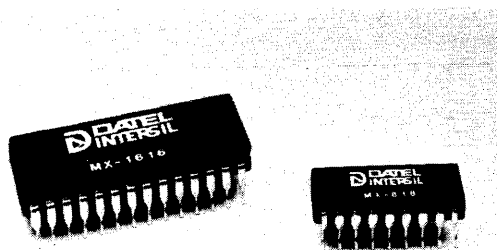
**SHM-20, Fast, High Performance Sample and Hold Features Internal Hold Capacitor.**

The SHM-20 is a complete monolithic sample and hold amplifier which includes an internal 100 pF MOS capacitor. Features include an acquisition time of typically 1.0 µS for a 10V input step to 0.01%. Aperture uncertainty is typically 1 nS and droop rate is as low as 0.08 µV/µS. Primarily designed for high speed analog signal processing applications, the SHM-20 combines monolithic reliability and size with high performance versatility and low cost. The SHM-20 is available for operation over the commercial, 0°C to +70°C and military, -55°C to +125°C temperature ranges and is packaged in a 14 pin ceramic DIP.



**SHM-4860 new 0.01% Hybrid Sample-Hold features 200 nsec acquisition time**

DATel's SHM-4860 is a high-speed, high resolution sample-hold amplifier manufactured with modern hybrid technology. Designed mainly for precision, high speed analog signal processing applications, the SHM-4860 acquires a 10V input change to ±0.01% in only 200 nsec max. Maximum sample-to-hold settling time is 100 nsec to ±0.01% with an aperture uncertainty of typically ±50 psec. Feedthrough attenuation is typically 74 dB. The digital inputs of the SHM-4860 are TTL compatible.



**8 and 16 Channel Multiplexers Feature High Speed, High Accuracy, MX-1616, MX-818**

DATel's MX-1616 and MX-818 are high speed, high performance analog multiplexers featuring transfer accuracies of 0.01% at sampling rates of up to 2.5 MHz over ±10V single ranges. The MX-1616 is user programmable either as a single-ended 16-channel or as a differential 8 channel multiplexer while the MX-818 is programmable as either a single-ended 8 channel or differential 4-channel multiplexer. All models feature an inhibit function, break-before-make switching and a maximum ON resistance of only 750Ω.



# Hybrid military products

Datel is a recognized industry leader in the design and manufacture of thin film hybrid data conversion products which meet the most demanding reliability requirements for military and aerospace applications per MIL-specifications. Datel data conversion products are currently used in a wide number of military and aerospace flight systems and in high reliability ground support and test systems. Datel's modern 120,000 square foot manufacturing facility in Mansfield, Massachusetts includes the most automated and advanced manufacturing, test and calibration equipment available in the industry. This capability, supported by a Quality Assurance Program with full emphasis on product quality assurance and reliability, provides an experienced and reliable source for data converter products to the screening and qualification requirements of Methods 5004, 5005, and 5008 of MIL-STD-883B in compliance with MIL-M-38510 (specified by Datel as "Q.L." devices).

The Quality Assurance operation at Datel monitors all areas of manufacturing and test, controls manufacturing and screening standards, maintains lot traceability procedures, and sets material standards to assure product quality. All purchased and internally manufactured components are procured or manufactured to precise specification control drawings. All components are 100% electrically tested and 100% visually inspected either by the vendors or internally within Datel facilities. Assembly and test processes and work stations are carefully monitored by Quality Assurance using fully documented procedures to guarantee high standards of workmanship and quality in all of Datel's products.

Datel products with the suffix "Q.L." are fully screened in accordance with Methods 5004 and 5008 of MIL-STD-883B as amended by MIL-M-38510. The following list briefly summarizes Datel's hybrid military products.

## ANALOG TO DIGITAL CONVERTERS

MODEL NO.	RESOLUTION	CONVERSION		LINEARITY	OPERATING TEMP. RANGE (°C)
		TIME			
ADC-HC12BMM	12 bits	300 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-HC12BMM-QL	12 bits	300 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-HS12BMM	12 bits	9 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-HS12BMM-QL	12 bits	9 $\mu$ sec		$\pm 1/2$ LSB	-55 to +100
ADC-HX12BMM	12 bits	20 $\mu$ sec		$\pm 1/2$ LSB	-55 to +100
ADC-HX12BMM-QL	12 bits	20 $\mu$ sec		$\pm 1/2$ LSB	-55 to +100
ADC-HZ12BMM	12 bits	8 $\mu$ sec		$\pm 1/2$ LSB	-55 to +100
ADC-HZ12BMM-QL	12 bits	8 $\mu$ sec		$\pm 1/2$ LSB	-55 to +100
ADC-810MM	12 bits	2 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-810MM-QL	12 bits	2 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-811MM	12 bits	4 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-811MM-QL	12 bits	4 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-815MM	8 bits	700 nsec		$\pm 1/2$ LSB	-55 to +125
ADC-815MM-QL	8 bits	700 nsec		$\pm 1/2$ LSB	-55 to +125
ADC-816MM	10 bits	800 nsec		$\pm 1/2$ LSB	-55 to +125
ADC-816MM-QL	10 bits	800 nsec		$\pm 1/2$ LSB	-55 to +125
ADC-817MM	12 bits	2 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-817MM-QL	12 bits	2 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-825MM	8 bits	1 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-825MM-QL	8 bits	1 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-826MM	10 bits	1.4 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-826MM-QL	10 bits	1.4 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-827MM	12 bits	3 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-827MM-QL	12 bits	3 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5101H	8 bits	900 nsec		$\pm 1/2$ LSB	-55 to +125
ADC-5101H-QL	8 bits	900 nsec		$\pm 1/2$ LSB	-55 to +125
ADC-5210H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5210H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5211H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5211H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5212H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5212H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5213H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5213H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5214H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5214H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5215H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5215H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5216H	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
ADC-5216H-QL	12 bits	13 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125

## DIGITAL-TO-ANALOG CONVERTERS

MODEL NO.	RESOLUTION	OUTPUT		LINEARITY	OPERATING TEMP. RANGE (°C)
		SETTLING TIME			
DAC-HA10BM	10 bits	1.3 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
DAC-HA10BM-QL	10 bits	1.3 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
DAC-HA12BM	12 bits	5 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
DAC-HA12BM-QL	12 bits	5 $\mu$ sec		$\pm 1/2$ LSB	-55 to +125
DAC-HA14BM	14 bits	7 $\mu$ sec		$\pm 1$ LSB	-55 to +125
DAC-HA14BM-QL	14 bits	7 $\mu$ sec		$\pm 1$ LSB	-55 to +125
DAC-HF8BMM	8 bits	25 nsec		$\pm 1/2$ LSB	-55 to +125
DAC-HF8BMM-QL	8 bits	25 nsec		$\pm 1/2$ LSB	-55 to +125
DAC-HF10BMM	10 bits	25 nsec		$\pm 1/2$ LSB	-55 to +125
DAC-HF10BMM-QL	10 bits	25 nsec		$\pm 1/2$ LSB	-55 to +125

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## SAMPLE-HOLD AMPLIFIERS

MODEL NO.	ACCURACY	ACQUISITION TIME	HOLD-MODE DROOP	OPERATING TEMP. RANGE (°C)
SHM-IC-1	0.01%	5 $\mu$ sec	50 $\mu$ V/msec	0 to + 70
SHM-IC-1M	0.01%	5 $\mu$ sec	50 $\mu$ V/msec	- 55 to + 125
SHM-LM-2	0.01%	6 $\mu$ sec	200 $\mu$ V/msec	0 to + 70
SHM-LM-2M	0.01%	6 $\mu$ sec	200 $\mu$ V/msec	- 55 to + 125
SHM-HUMC	0.1%	25 nsec	50 $\mu$ V/ $\mu$ sec	0 to + 70
SHM-HUMR	0.1%	25 nsec	50 $\mu$ V/ $\mu$ sec	- 25 to + 85
SHM-6MC	0.01%	1 $\mu$ sec	10 $\mu$ V/ $\mu$ sec	0 to + 70
SHM-6MR	0.01%	1 $\mu$ sec	10 $\mu$ V/ $\mu$ sec	- 25 to + 85
SHM-7MC	0.1%	40 nsec	100 $\mu$ V/ $\mu$ sec	0 to + 70
SHM-7MR	0.1%	40 nsec	100 $\mu$ V/ $\mu$ sec	- 25 to + 85
NEW SHM-9MC	0.01%	6 $\mu$ sec	0.2 mV/mS	0 to + 70
NEW SHM-9MR	0.01%	6 $\mu$ sec	0.2 mV/mS	- 25 to + 85
NEW SHM-20C	0.01%	1 $\mu$ sec	0.08 $\mu$ V/ $\mu$ S	0 to + 70
NEW SHM-20M	0.01%	1 $\mu$ sec	0.08 $\mu$ V/ $\mu$ S	- 55 to + 125
NEW SHM-4860MC	0.01%	200 nsec	0.5 $\mu$ V/ $\mu$ S	0 to + 70
NEW SHM-4860MR	0.01%	200 nsec	0.5 $\mu$ V/ $\mu$ S	- 25 to + 85

## ANALOG MULTIPLEXERS

MODEL NO.	INPUT CHANNELS	ACCESS TIME	CHANNEL ON RESISTANCE	OPERATING TEMP. RANGE (°C)
MV-808	8 Single-Ended	350 nsec	250 $\Omega$	0 to + 70
MV-808M	8 Single-Ended	350 nsec	250 $\Omega$	- 55 to + 125
MV-1606	16 Single-Ended	300 nsec	270 $\Omega$	0 to + 70
MV-1606M	16 Single-Ended	300 nsec	270 $\Omega$	- 55 to + 125
MVD-409	4 Differential	350 nsec	250 $\Omega$	0 to + 70
MVD-409M	4 Differential	350 nsec	250 $\Omega$	- 55 to + 125
MVD-807	8 Differential	300 nsec	270 $\Omega$	0 to + 70
MVD-807M	8 Differential	300 nsec	270 $\Omega$	- 55 to + 125
MX-808	8 Single-Ended	500 nsec	1.5k $\Omega$	0 to + 70
MX-808M	8 Single-Ended	500 nsec	1.5k $\Omega$	- 55 to + 125
MX-818C	8 Single-Ended or 4 Differential	125 nsec	750 $\Omega$	0 to + 70
MX-818M	8 Single-Ended or 4 Differential	125 nsec	750 $\Omega$	- 55 to + 125
MX-1606	16 Single-Ended	500 nsec	1.5k $\Omega$	0 to + 70
MX-1606M	16 Single-Ended	500 nsec	1.5k $\Omega$	- 55 to + 125
MX-1616C	16 Single-Ended 8 Differential	150 nsec	750 $\Omega$	0 to + 70
MX-1616M	16 Single-Ended 8 Differential	150 nsec	750 $\Omega$	- 55 to + 125
MXD-409	4 Differential	500 nsec	1.5k $\Omega$	0 to + 70
MXD-409M	4 Differential	500 nsec	1.5k $\Omega$	- 55 to + 125
MXD-807	8 Differential	500 nsec	1.5k $\Omega$	0 to + 70
MXD-807M	8 Differential	500 nsec	1.5k $\Omega$	- 55 to + 125

## OPERATIONAL AMPLIFIERS

MODEL NO.	INPUT OFFSET VOLTAGE	GAIN BANDWIDTH	OUTPUT	OPERATING TEMP. RANGE (°C)
AM-410-2C	1.5 mV	18 MHz	$\pm$ 11V @ 8 mA	0 to + 70
AM-410-2M	1 mV	18 MHz	$\pm$ 12V @ 10 mA	- 55 to + 125
AM-411-2C	1.5 mV	50 MHz	$\pm$ 11V @ 8 mA	0 to + 70
AM-411-2M	1 mV	60 MHz	$\pm$ 12V @ 10 mA	- 55 to + 125
AM-427-A	100 $\mu$ V	5 MHz	$\pm$ 11V @ 18 mA	- 25 to + 85
AM-427-B	25 $\mu$ V	5 MHz	$\pm$ 11V @ 18 mA	- 25 to + 85
AM-427-M	100 $\mu$ V	5 MHz	$\pm$ 11V @ 18 mA	- 55 to + 125
AM-430A	75 $\mu$ V	2.5 MHz	$\pm$ 10V @ 18 mA	0 to + 70
AM-430B	25 $\mu$ V	2.5 MHz	$\pm$ 10V @ 18 mA	0 to + 70
AM-430M	75 $\mu$ V	2.5 MHz	$\pm$ 10V @ 25 mA	- 55 to + 125
AM-450-2	4 mV	12 MHz	$\pm$ 10V @ 10 mA	0 to + 70
AM-450-2M	4 mV	12 MHz	$\pm$ 10V @ 10 mA	- 55 to + 125
AM-452-2	5 mV	20 MHz	$\pm$ 10V @ 10 mA	0 to + 70
AM-452-2M	5 mV	20 MHz	$\pm$ 10V @ 10 mA	- 55 to + 125
AM-453-2	4 mV	10 MHz	$\pm$ 12V @ 20 mA	- 55 to + 125
AM-453-2M	4 mV	10 MHz	$\pm$ 12V @ 20 mA	- 55 to + 125
AM-460-2	3 mV	12 MHz	$\pm$ 10V @ 10 mA	0 to + 70
AM-460-2M	3 mV	12 MHz	$\pm$ 10V @ 10 mA	- 55 to + 125
AM-462-2	3 mV	100 MHz	$\pm$ 10V @ 10 mA	0 to + 70
AM-462-2M	3 mV	100 MHz	$\pm$ 10V @ 10 mA	- 55 to + 125
AM-464-2	6 mV	4 MHz	$\pm$ 35V @ 10 mA	0 to + 70

**OPERATIONAL AMPLIFIERS**

MODEL NO.	INPUT OFFSET VOLTAGE	GAIN BANDWIDTH	OUTPUT	OPERATING TEMP. RANGE (°C)
AM-464-2M	4 mV	4 MHz	± 35V @ 12 mA	-55 to + 125
AM-470-2C	5 mV	1 MHz	± 12V @ 10 mA	0 to + 70
AM-470-2M	3 mV	1 MHz	± 12V @ 10 mA	-55 to + 125
AM-490-2A	20 μV	3 MHz	± 10V @ 7 mA	0 to + 70
AM-490-2B	20 μV	3 MHz	± 10V @ 7 mA	0 to + 70
AM-490-2C	20 μV	3 MHz	± 10V @ 7 mA	0 to + 70
AM-490-2M	20 μV	3 MHz	± 10V @ 7 mA	-55 to + 125
AM-500GC	3 mV	130 MHz	± 10V @ 50 mA	0 to + 70
AM-500MC	3 mV	130 MHz	± 10V @ 50 mA	0 to + 70
AM-500MR	3 mV	130 MHz	± 10V @ 50 mA	-25 to + 85
NEW AM-1435MC	5 mV	1000 MHz	± 7V @ 14 mA	0 to + 70
NEW AM-1435MR	5 mV	1000 MHz	± 7V @ 14 mA	-25 to + 85
NEW AM-7650-1	5 μV	2 MHz	± 4.7V	0 to + 70
NEW AM-7650-2	5 μV	2 MHz	± 4.7V	0 to + 70

**DIGITALLY PROGRAMMABLE INSTRUMENTATION AMPLIFIERS**

MODEL NO.	GAIN RANGE	SETTLING TIME	OUTPUT	OPERATING TEMP. RANGE (°C)
AM-542MC	1 to 1024	150 μsec	± 10.5V @ 5 mA	0 to + 70
AM-542MR	1 to 1024	150 μsec	± 10.5V @ 5 mA	-25 to + 85
AM-543MC	1 to 128	6 μsec	± 11V @ 1 mA	0 to + 70
AM-543MR	1 to 128	6 μsec	± 11V @ 1 mA	-25 to + 85

**PROGRAMMABLE GAIN INSTRUMENTATION AMPLIFIERS**

MODEL NO.	GAIN RANGE	SETTLING TIME	OUTPUT	OPERATING TEMP. RANGE (°C)
NEW AM-551MC	1 to 1000	2 μsec	± 11V @ 5 mA	0 to + 70
NEW AM-551MR	1 to 1000	2 μsec	± 11V @ 5 mA	-25 to + 85
NEW AM-551MC	1 to 1000	2 μsec	± 11V @ 5 mA	0 to + 70

**DATA ACQUISITION SUBSYSTEMS**

MODEL NO.	RESOLUTION	INPUT CHANNELS	THROUGHOUT RATE	OPERATING TEMP. RANGE (°C)
DAS-952R	8 bits	16 Single-Ended	17 kHz	-25 to + 85
HDAS-8MC	12 bits	8 Differential	50 kHz	0 to + 70
HDAS-8MR	12 bits	8 Differential	50 kHz	-25 to + 85
HDAS-16MC	12 bits	16 Single-Ended	50 kHz	0 to + 70
HDAS-16MR	12 bits	16 Single-Ended	50 kHz	-25 to + 85

**VOLTAGE TO FREQUENCY CONVERTERS**

MODEL NO.	LINEARITY	OUTPUT RANGE	GAIN TEMPCO	OPERATING TEMP. RANGE (°C)
VFQ-1C	0.05%	10 kHz to 100 kHz	40 ppm/°C	0 to + 70
VFQ-1R	0.05%	10 kHz to 100 kHz	40 ppm/°C	-25 to + 85
VFQ-2C	0.01%	10 kHz to 100 kHz	40 ppm/°C	0 to + 70
VFQ-3C	0.25%	10 kHz to 100 kHz	40 ppm/°C	0 to + 70

**ACTIVE FILTERS**

MODEL NO.	FREQUENCY RANGE	F <sub>0</sub> ACCURACY	Q RANGE	OPERATING TEMP. RANGE (°C)
FLT-U2	0.01 Hz to 200 kHz	± 5%	0.1 to 1000	0 to + 70
FLT-U2M	0.01 Hz to 200 kHz	± 5%	0.1 to 1000	-55 to + 125

**MONOLITHIC VOLTAGE CONVERTER**

MODEL NO.	INPUT VOLTAGE RANGE	INPUT CURRENT	OUTPUT VOLTAGE RANGE	OPERATING TEMP. RANGE (°C)
NEW VI-7660	+1.5V to +10.0V	500 μA	-1.5V to -10.0V	0 to + 70

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• ANALOG VOLTAGE RANGE <span style="float: right;">±15V</span></li> <li>• ANALOG CURRENT RANGE <span style="float: right;">80mA</span></li> <li>• TURN-ON TIME <span style="float: right;">240ns</span></li> <li>• LOW R<sub>ON</sub> <span style="float: right;">55Ω</span></li> <li>• LOW POWER DISSIPATION <span style="float: right;">15mW</span></li> <li>• TTL/CMOS COMPATIBLE</li> </ul>	<p>HI-200 is a monolithic device comprising two independently selectable SPST switches which feature fast switching speeds (290ns) combined with low power dissipation (15mW at 25°C). Each switch provides low "ON" resistance operation for input signal voltages up to the supply rails and for signal currents up to 80mA. Employing Dielectric Isolation and CMOS processing, HI-200 operates without any applications problems induced by latch-up or SCR mode phenomena.</p>
<h3>APPLICATIONS</h3>	<p>All devices provide break-before-make switching and are TTL and CMOS compatible for maximum application versatility. HI-200 is an ideal component for use in high frequency analog switching. Typical applications include signal path switching, sample and hold circuit, digital filters, and op amp gain switching networks.</p>
<ul style="list-style-type: none"> <li>• HIGH FREQUENCY ANALOG SWITCHING</li> <li>• SAMPLE AND HOLD CIRCUITS</li> <li>• DIGITAL FILTERS</li> <li>• OP AMP GAIN SWITCHING NETWORKS</li> </ul>	<p>HI-200 is available in DIP and metal (TO-100) cans. HI-200-2 is specified from -55°C to +125°C while HI-200-5 operates from 0°C to +75°C. HI-200 is functionally and pin compatible with other available "200 series" switches.</p>
<h3>PINOUT</h3>	<h3>FUNCTIONAL DIAGRAM</h3>
<p style="text-align: center;">Section 11 for Packaging</p> <div style="display: flex; justify-content: space-around;"> <div data-bbox="97 1449 422 1701"> </div> <div data-bbox="438 1386 730 1722"> <p style="text-align: center;">Top View</p> </div> </div>	<p style="text-align: center;">SWITCH OPEN FOR LOGIC HIGH</p>

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage	44V (±22)	Total Power Dissipation*	450mW
V <sub>REF</sub> to Ground	+20V, -5V	Operating Temperature	
Digital Input Voltage:	+V <sub>Supply</sub> +4V	HI-200-2	-55°C to +125°C
	-V <sub>Supply</sub> -4V	HI-200-4	-20°C to +85°C
Analog Input Voltage (One Switch)	+V <sub>Supply</sub> +2.0V	HI-200-5	0°C to +75°C
	-V <sub>Supply</sub> -2.0V	Storage Temperature	-65°C to +150°C

\*Derate 6mW/°C Above T<sub>A</sub> = 75°C

## ELECTRICAL CHARACTERISTICS

Unless Otherwise Specified

Supplies = +15V, -15V; V<sub>REF</sub> = Open; V<sub>AH</sub>(Logic Level High) = 2.4V V<sub>AL</sub>(Logic Level Low) = +0.8V

For Test Conditions, consult Performance Characteristics

PARAMETER	TEMP.	HI-200-2 -55°C to +125°C			HI-200-5 ** 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b><u>ANALOG SWITCH CHARACTERISTICS</u></b>								
V <sub>S</sub> , Analog Signal Range	Full	-15		+15	-15		+15	V
R <sub>ON</sub> , On Resistance (Note 1)	+25°C Full		55 80	70 100		55 72	80 100	Ω Ω
I <sub>S</sub> (OFF), Off Input Leakage Current (Note 6)	+25°C Full		1 100	500		1 10	500	nA nA
I <sub>D</sub> (OFF), Off Output Leakage Current (Note 6)	+25°C Full		1 100	500		1 10	500	nA nA
I <sub>D</sub> (ON), On Leakage Current (Note 6)	+25°C Full		.02 6	500		.02 6	500	nA nA
<b><u>DIGITAL INPUT CHARACTERISTICS</u></b>								
V <sub>AL</sub> , Input Low Threshold	Full			0.8			0.8	V
V <sub>AH</sub> , Input High Threshold	Full	2.4			2.4			V
I <sub>A</sub> , Input Leakage Current (High or Low) (Note 2)	Full			1.0			1.0	μA
<b><u>SWITCHING CHARACTERISTICS</u></b>								
t <sub>OPEN</sub> , Break - Before Make Delay (Note 3)	+25°C		60			60		ns
t <sub>ON</sub> , Switch on Time	+25°C		240	500		240		ns
t <sub>OFF</sub> , Switch off Time	+25°C		330	500		500		ns
"Off Isolation" (Note 4)	+25°C		70			70		dB
C <sub>S</sub> (OFF), Input Switch Capacitance	+25°C		5.5			5.5		pF
C <sub>D</sub> (OFF), { C <sub>D</sub> (ON), } Output Switch Capacitance	+25°C		5.5			5.5		pF
	+25°C		11			11		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		5			5		pF
C <sub>DS</sub> (OFF), Drain-To-Source Capacitance	+25°C		0.5			0.5		pF
<b><u>POWER REQUIREMENTS</u> (Note 5)</b>								
P <sub>D</sub> , Power Dissipation	+25°C Full		15	60		15	60	mW mW
I <sup>+</sup> , Current	+25°C Full		0.5	2.0		0.5	2.0	mA mA
I <sup>-</sup> , Current	+25°C Full		0.5	2.0		0.5	2.0	mA mA

- NOTES:
- V<sub>OUT</sub> = ±10V I<sub>OUT</sub> = 1mA
  - Digital Inputs Are MOS Gates - Typical Leakage is Less Than 1nA
  - V<sub>AH</sub> = 4.0V

- V<sub>A</sub> = +5V, R<sub>L</sub> = 1KΩ, C<sub>L</sub> = 10pF, V<sub>S</sub> = 3VRMS, f = 100kHz
- V<sub>A</sub> = +3V or V<sub>A</sub> = 0V For Both Switches
- Refer to leakage current measurement diagram on page (3-8)

\*\* Note: HI-200-4 has same specifications as HI-200-5 over the temperature range -20°C to +85°C.



# HARRIS

# HI-201

## Quad SPST CMOS Analog Switch

### FEATURES

- ANALOG VOLTAGE RANGE  $\pm 15V$
- ANALOG CURRENT RANGE 80mA
- TURN-ON TIME 185ns
- LOW  $R_{ON}$  55  $\Omega$
- LOW POWER DISSIPATION 15mW
- TTL/CMOS COMPATIBLE

### DESCRIPTION

HI-201 is a monolithic device comprising four independently selectable SPST switches which feature fast switching speeds (185ns) combined with low power dissipation (15mW at 25°C). Each switch provides low "ON" resistance operation for input signal voltages up to the supply rails and for signal currents up to 80mA. Employing Dielectric Isolation and CMOS processing, HI-201 operates without any applications problems induced by latch-up or SCR-mode phenomena.

### APPLICATIONS

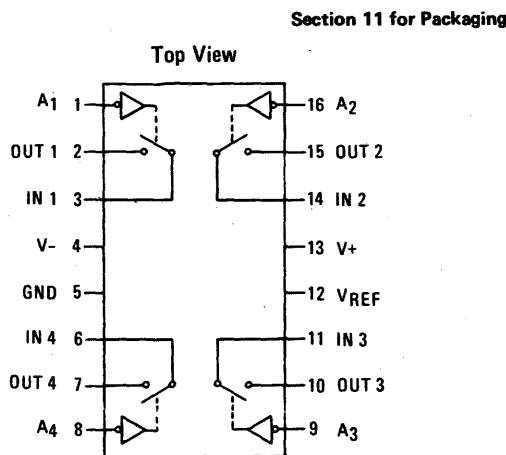
- HIGH FREQUENCY ANALOG SWITCHING
- SAMPLE AND HOLD CIRCUITS
- DIGITAL FILTERS
- OP AMP GAIN SWITCHING NETWORKS

All devices provide break-before-make switching and are TTL and CMOS compatible for maximum application versatility. HI-201 is an ideal component for use in high frequency analog switching. Typical applications include signal path switching, sample and hold circuit, digital filters, and op amp gain switching networks.

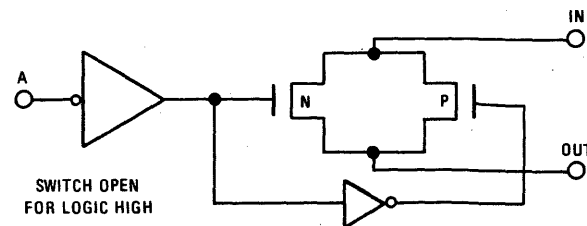
HI-201 is available in a 16 lead dual-in-line package. HI-201-2 is specified from -55°C to +125°C while HI-201-5 operates from 0°C to +75°C. HI-201 is functionally and pin compatible with other available "200 series" switches.

### PIN OUT

### FUNCTIONAL DIAGRAM



TYPICAL SWITCH



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage	44V (±22)	Total Power Dissipation*	450mW
VREF to Ground	+20V, -5V	Operating Temperature	
Digital Input Voltage:	+VSupply +4V	HI-200-2	-55°C to +125°C
	-VSupply -4V	HI-200-4	-20°C to +85°C
Analog Input Voltage (One Switch)	+VSupply +2.0V	HI-200-5	0°C to +75°C
	-VSupply -2.0V	Storage Temperature	-65°C to +150°C

\*Derate 6mW/°C Above TA = 75°C

**ELECTRICAL CHARACTERISTICS**

Unless Otherwise Specified

Supplies = +15V, -15V; VREF = Open; VAH (Logic Level High) = 2.4V VAL (Logic Level Low) = +0.8V

For Test Conditions, consult Performance Characteristics

PARAMETER	TEMP.	HI-200-2 -55°C to +125°C			HI-200-5 ** 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b><u>ANALOG SWITCH CHARACTERISTICS</u></b>								
V <sub>S</sub> , Analog Signal Range	Full	-15		+15	-15		+15	V
R <sub>ON</sub> , On Resistance (Note 1)	+25°C		55	70		55	80	Ω
	Full		80	100		72	100	Ω
I <sub>S</sub> (OFF), Off Input Leakage Current (Note 6)	+25°C		1			1		nA
	Full		100	500		10	500	nA
I <sub>D</sub> (OFF), Off Output Leakage Current (Note 6)	+25°C		1			1		nA
	Full		100	500		10	500	nA
I <sub>D</sub> (ON), On Leakage Current (Note 6)	+25°C		.02			.02		nA
	Full		6	500		6	500	nA
<b><u>DIGITAL INPUT CHARACTERISTICS</u></b>								
VAL, Input Low Threshold	Full			0.8			0.8	V
VAH, Input High Threshold	Full	2.4			2.4			V
I <sub>A</sub> , Input Leakage Current (High or Low) (Note 2)	Full			1.0			1.0	μA
<b><u>SWITCHING CHARACTERISTICS</u></b>								
t <sub>OPEN</sub> , Break - Before Make Delay (Note 3)	+25°C		60			60		ns
t <sub>on</sub> , Switch on Time	+25°C		240	500		240		ns
t <sub>off</sub> , Switch off Time	+25°C		330	500		500		ns
"Off Isolation" (Note 4)	+25°C		70			70		dB
C <sub>S</sub> (OFF), Input Switch Capacitance	+25°C		5.5			5.5		pF
C <sub>D</sub> (OFF), { C <sub>D</sub> (ON), } Output Switch Capacitance	+25°C		5.5			5.5		pF
	+25°C		11			11		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		5			5		pF
C <sub>DS</sub> (OFF), Drain-To-Source Capacitance	+25°C		0.5			0.5		pF
<b><u>POWER REQUIREMENTS</u> (Note 5)</b>								
P <sub>D</sub> , Power Dissipation	+25°C		15			15		mW
I <sup>+</sup> , Current	Full			60			60	mW
	+25°C		0.5			0.5		mA
I <sup>-</sup> , Current	Full			2.0			2.0	mA
	+25°C		0.5			0.5		mA
	Full			2.0			2.0	mA

- NOTES:
- V<sub>OUT</sub> = ±10V I<sub>OUT</sub> = 1mA
  - Digital Inputs Are MOS Gates - Typical Leakage is Less Than 1nA
  - V<sub>AH</sub> = 4.0V

- V<sub>A</sub> = +5V, R<sub>L</sub> = 1KΩ, C<sub>L</sub> = 10pF, V<sub>S</sub> = 3VRMS, f = 100kHz
- V<sub>A</sub> = +3V or V<sub>A</sub> = 0V For Both Switches
- Refer to leakage current measurement diagram on page (3-8)

\*\* Note: HI-200-4 has same specifications as HI-200-5 over the temperature range -20°C to +85°C.



# HI-201

## Quad SPST CMOS Analog Switch

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• ANALOG VOLTAGE RANGE <span style="float: right;"><math>\pm 15V</math></span></li> <li>• ANALOG CURRENT RANGE <span style="float: right;">80mA</span></li> <li>• TURN-ON TIME <span style="float: right;">185ns</span></li> <li>• LOW <math>R_{ON}</math> <span style="float: right;"><math>55\ \Omega</math></span></li> <li>• LOW POWER DISSIPATION <span style="float: right;">15mW</span></li> <li>• TTL/CMOS COMPATIBLE</li> </ul>	<p>HI-201 is a monolithic device comprising four independently selectable SPST switches which feature fast switching speeds (185ns) combined with low power dissipation (15mW at 25°C). Each switch provides low "ON" resistance operation for input signal voltages up to the supply rails and for signal currents up to 80mA. Employing Dielectric Isolation and CMOS processing, HI-201 operates without any applications problems induced by latch-up or SCR-mode phenomena.</p> <p>All devices provide break-before-make switching and are TTL and CMOS compatible for maximum application versatility. HI-201 is an ideal component for use in high frequency analog switching. Typical applications include signal path switching, sample and hold circuit, digital filters, and op amp gain switching networks.</p> <p>HI-201 is available in a 16 lead dual-in-line package. HI-201-2 is specified from -55°C to +125°C while HI-201-5 operates from 0°C to +75°C. HI-201 is functionally and pin compatible with other available "200 series" switches.</p>
APPLICATIONS	
PIN OUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;">Section 11 for Packaging</p> <p style="text-align: center;">Top View</p>	<p>TYPICAL SWITCH</p>



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage Between Pins 4 and 13	44V (±22)	Total Power Dissipation*	750mW
V <sub>REF</sub> to Ground	+20V, -5V	Operating Temperature	
Digital Input Voltage:	V <sub>Supply(+)</sub> +4V	HI-201-2	-55°C to +125°C
	V <sub>Supply(-)</sub> -4V	HI-201-4	-20°C to +85°C
Analog Input Voltage (One Switch)	+V <sub>Supply</sub> +2.0V	HI-201-5	0°C to +75°C
	-V <sub>Supply</sub> -2.0V	Storage Temperature	-65°C to +150°C

\*Derate 8mW/°C Above T<sub>A</sub> = +75°C

## ELECTRICAL CHARACTERISTICS

Unless Otherwise Specified

Supplies = +15V, -15V; V<sub>REF</sub> = Open; V<sub>AH</sub> (Logic Level High) = 2.4V V<sub>AL</sub> (Logic Level Low) = +0.8V

For Test Conditions, consult Performance Characteristics

PARAMETER	TEMP.	HI-201-2			HI-201-5 **			UNITS
		-55°C to +125°C			0°C to +75°C			
<b>ANALOG SWITCH CHARACTERISTICS</b>								
V <sub>S</sub> , Analog Signal Range	Full	-15		+15	-15		+15	V
R <sub>ON</sub> , On Resistance (Note 1)	+25°C		55	70		55	80	Ω
	Full		80	100		75	100	Ω
I <sub>S(OFF)</sub> , Off Input Leakage Current (Note 6)	+25°C		2			2		nA
	Full			500			250	nA
I <sub>D(OFF)</sub> , Off Output Leakage Current (Note 6)	+25°C		2			2		nA
	Full			500			250	nA
I <sub>D(ON)</sub> , On Leakage Current (Note 6)	+25°C		2			2		nA
	Full			500			250	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>AL</sub> , Input Low Threshold	Full			0.8			0.8	V
V <sub>AH</sub> , Input High Threshold	Full	2.4			2.4			V
I <sub>A</sub> , Input Leakage Current (High or Low) (Note 2)	Full			1.0			1.0	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>OPEN</sub> , Break - Before Make Delay (Note 3)	+25°C		30			30		ns
t <sub>on</sub> , Switch ON Time	+25°C		185	500		185		ns
t <sub>off</sub> , Switch OFF Time	+25°C		220	500		220		ns
"Off Isolation" (Note 4)	+25°C		80			80		dB
C <sub>S(OFF)</sub> , Input Switch Capacitance	+25°C		5.5			5.5		pF
C <sub>D(OFF)</sub> , { Output Switch Capacitance	+25°C		5.5			5.5		pF
	+25°C		11			11		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		5			5		pF
C <sub>DS(OFF)</sub> , Drain-To-Source Capacitance	+25°C		0.5			0.5		pF
<b>POWER REQUIREMENTS (Note 5)</b>								
P <sub>D</sub> , Power Dissipation	+25°C		15			15		mW
	Full			60			60	mW
I <sub>+</sub> , Current (Pin 13)	+25°C		0.5			0.5		mA
	Full			2.0			2.0	mA
I <sub>-</sub> , Current (Pin 4)	+25°C		0.5			0.5		mA
	Full			2.0			2.0	mA

- NOTES: 1. V<sub>OUT</sub> = +10V I<sub>OUT</sub> = 1mA  
 2. Digital Inputs Are MOS Gates - Typical Leakage is Less Than 1nA  
 3. V<sub>AH</sub> = 4.0V

4. V<sub>A</sub> = 5V, R<sub>L</sub> = 1KΩ, C<sub>L</sub> = 10pF, V<sub>S</sub> = 3VRMS, f = 100KHz  
 5. V<sub>A</sub> = +3V or V<sub>A</sub> = 0V For all Switches  
 6. Refer to leakage current measurement diagram on page (3-14)

\*\* Note: HI-201-4 has same specifications as HI-201-5 over the temperature range -20°C to +85°C.



# HI-201HS

## High Speed Quad SPST CMOS Analog Switch

*Preliminary*

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• ANALOG VOLTAGE RANGE <span style="float: right;">±15V</span></li> <li>• ANALOG CURRENT RANGE <span style="float: right;">80mA</span></li> <li>• TURN-ON TIME <span style="float: right;">30ns</span></li> <li>• LOW R<sub>ON</sub> <span style="float: right;">30Ω</span></li> <li>• LOW POWER DISSIPATION <span style="float: right;">120mW</span></li> <li>• TTL COMPATIBLE</li> <li>• LOW CHARGE INJECTION <span style="float: right;">10pC</span></li> </ul>	<p>The Harris HI-201HS is a monolithic CMOS analog switch featuring very fast switching speeds and low ON resistance. The device consists of four independently selectable SPST switches and is identical in pinout to the HI-201 quad switch.</p> <p>Fabricated using the Harris dielectric isolation technology, this TTL compatible device offers improved performance over previously available CMOS analog switches. Featuring switching speeds of 50ns max., low ON resistance of 50Ω max., and wide analog signal range of ±15V, the HI-201HS is designed for any application where improved switching performance, particularly switching speed, is required.</p> <p>The HI-201HS is available in a 16 lead dual-in-line package. The HI-201HS-2 is specified for the temperature range of -55°C to +125°C and the HI-201HS-5 operates from 0°C to +75°C.</p>
<h3>APPLICATIONS</h3> <ul style="list-style-type: none"> <li>• HIGH FREQUENCY ANALOG SWITCHING</li> <li>• SAMPLE AND HOLD CIRCUITS</li> <li>• DIGITAL FILTERS</li> <li>• OP AMP GAIN SWITCHING NETWORKS</li> </ul>	
PIN OUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;">Top View</p>	<p>TYPICAL SWITCH</p>



**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Supply Voltage Between Pins 4 and 13	+36V	Total Power Dissipation*	750mW
Digital Input Voltage:	V <sub>Supply</sub> (+) +4V	Operating Temperature	
	V <sub>Supply</sub> (-) -4V	HI-201HS-2	-55°C to +125°C
Analog Input Voltage (One Switch)	+V <sub>Supply</sub> +2.0V	HI-201HS-4	-20°C to +85°C
	-V <sub>Supply</sub> -2.0V	HI-201HS-5	0°C to +75°C
		Storage Temperature	-65°C to +150°C

\*Derate 8mW/°C Above T<sub>A</sub> = +75°C

**ELECTRICAL CHARACTERISTICS** Unless Otherwise Specified, Supplies = +15V, -15V; V<sub>AH</sub> (Logic Level High) = 5.0V; V<sub>AL</sub> (Logic Level Low) = +0.8V

PARAMETER	TEMP.	HI-201HS-2 HI-201HS-5			UNITS
		MIN.	TYP.	MAX.	
<b>ANALOG SWITCH CHARACTERISTICS</b>					
V <sub>S</sub> , Analog Signal Range	Full	-15		+15	V
R <sub>ON</sub> , On Resistance (Note 2)	+25°C Full		30 75	50	Ω Ω
I <sub>S(OFF)</sub> , Off Input Leakage Current	+25°C Full		.3 100	10	nA nA
I <sub>D(OFF)</sub> , Off Output Leakage Current	+25°C Full		.3 100	10	nA nA
I <sub>D(ON)</sub> , On Leakage Current	+25°C Full		.1 100	10	nA nA
<b>DIGITAL INPUT CHARACTERISTICS</b>					
V <sub>AL</sub> , Input Low Threshold	Full			0.8	V
V <sub>AH</sub> , Input High Threshold	+25°C Full	2.0 2.4			V V
I <sub>AL</sub> , Input Leakage Current (Low)	Full			500	μA
I <sub>AH</sub> , Input Leakage Current (High)	Full			40	μA
<b>SWITCHING CHARACTERISTICS</b>					
t <sub>ON</sub> , Switch ON Time (Note 3)	+25°C		30	50	ns
t <sub>OFF</sub> , Switch OFF Time (Note 3)	+25°C		40	50	ns
"Off Isolation" (Note 4)	+25°C		72		dB
Crosstalk (Note 5)	+25°C		86		dB
Charge Injection (Note 6)	+25°C		10		pC
C <sub>S(OFF)</sub> , Input Switch Capacitance	+25°C		10		pF
C <sub>D(OFF)</sub> , Output Switch Capacitance	+25°C		10		pF
C <sub>D(ON)</sub>	+25°C		30		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		18		pF
C <sub>DS(OFF)</sub> , Drain-to-Source Capacitance	+25°C		.5		pF
<b>POWER REQUIREMENTS (Note 7)</b>					
P <sub>D</sub> , Power Dissipation	+25°C Full		120	240	mW mW
I <sup>+</sup> , Current (Pin 13)	+25°C Full		4.5	10.0	mA mA
I <sup>-</sup> , Current (Pin 4)	+25°C Full		3.5	6	mA mA

**NOTES:**

- Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- V<sub>OUT</sub> = ±10V, I<sub>OUT</sub> = 1mA
- R<sub>L</sub> = 1kΩ, C<sub>L</sub> = 35pF, V<sub>IN</sub> = +10V, V<sub>A</sub> = +5V
- V<sub>A</sub> = 5V, R<sub>L</sub> = 1kΩ, C<sub>L</sub> = 10pF, V<sub>S</sub> = 3 VRMS, f = 100kHz
- V<sub>A</sub> = 5V, R<sub>L</sub> = 1kΩ, f = 100kHz, V<sub>IN</sub> = 2Vp-p
- C<sub>L</sub> = 1000pF, V<sub>IN</sub> = 0V, R<sub>IN</sub> = 0Ω  
ΔQ = C<sub>L</sub> × ΔV<sub>O</sub>
- V<sub>A</sub> = 5V or V<sub>A</sub> = 0 for all switches.



# HI-300 thru HI-307

## CMOS Analog Switches

FEATURES	APPLICATIONS
<ul style="list-style-type: none"> <li>ANALOG SIGNAL RANGE (<math>\pm 15V</math> SUPPLIES) <math>\pm 15V</math></li> <li>LOW LEAKAGE (TYP. @ 25°C) 40pA</li> <li>LOW LEAKAGE (TYP. @ 125°C) 1nA</li> <li>LOW ON RESISTANCE (TYP. @ 25°C) 35<math>\Omega</math></li> <li>BREAK-BEFORE-MAKE DELAY (TYP.) 60ns</li> <li>CHARGE INJECTION 30pC</li> <li>TTL, CMOS COMPATIBLE</li> <li>SYMMETRICAL SWITCH ELEMENTS</li> <li>LOW OPERATING POWER (TYP. FOR HI-300 - 303) 1.0mW</li> </ul>	<ul style="list-style-type: none"> <li>SAMPLE AND HOLD i.e. LOW LEAKAGE SWITCHING</li> <li>OP AMP GAIN SWITCHING i.e. LOW ON RESISTANCE</li> <li>PORTABLE, BATTERY OPERATED CIRCUITS</li> <li>LOW LEVEL SWITCHING CIRCUITS</li> <li>DUAL OR SINGLE SUPPLY SYSTEMS</li> </ul>

FUNCTIONAL DIAGRAM	DESCRIPTION
<p style="text-align: center;">TYPICAL SWITCH 300 SERIES</p>	<p>The HI-300 through HI-307 series of switches are monolithic devices fabricated using CMOS technology and the Harris dielectric isolation process. These switches feature break-before-make switching, (HI-301, 303, 305 &amp; 307 only), low and nearly constant ON resistance over the full analog signal range, and low power dissipation, (a few milliwatts for the HI-300-303, a few hundred microwatts for the HI-304-307).</p> <p>The HI-300-303 are TTL compatible and have a logic "0" condition with an input less than 0.8V and a logic "1" condition with an input greater than 4.0V. The HI-304-307 switches are CMOS compatible and have a low state with an input less than 3.5V and a high state with an input greater than 11V. (See pinouts for switch conditions with a logic "1" input.)</p> <p>All the devices are available in a 14 pin epoxy or ceramic DIP. The HI-300, 301, 304 and 305 are also available in a 10 pin metal can. Each of the switch types are available in either the -55°C to +125°C or 0°C to +75°C operating ranges.</p>

**PINOUTS** (SWITCH STATES ARE FOR A LOGIC "1" INPUT) Section 11 for Packaging

**DUAL SPST HI-300 & HI-304 (TOP VIEWS)**

LOGIC	SWITCH
0	OFF
1	ON

\*The substrate and case are internally tied to V-. (The case should not be used as the V- connection, however.)

**SPDT HI-301 & HI-305 (TOP VIEWS)**

LOGIC	SW 1	SW 2
0	OFF	ON
1	ON	OFF

\*The substrate and case are internally tied to V-. (The case should not be used as the V- connection, however.)

**DUAL DPST HI-302 & HI-306 (TOP VIEW)**

LOGIC	SWITCH
0	OFF
1	ON

**DUAL SPDT HI-303 & HI-307 (TOP VIEW)**

LOGIC	SW 1	SW 2	SW 3	SW 4
0	OFF	ON	ON	OFF
1	ON	OFF	OFF	ON

# SPECIFICATIONS HI-300 - HI-307



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Voltage Between Supplies	44V (±22V)	Total Power Dissipation	
Digital Input Voltage	V <sup>+</sup> +4.0V V <sup>-</sup> -4.0V	14 Pin Epoxy DIP	526mW
		14 Pin Ceramic DIP	588mW
		10 Pin Metal Can*	435mW
		*Derate 6.9mW/0°C Above T <sub>A</sub> = 70°C	
Analog Input Voltage	V <sup>+</sup> 1.5V V <sup>-</sup> 1.5V	Operating Temperature	HI-3XX-2 -55°C to +125°C HI-3XX-5 0°C to +75°C
		Storage Temperature	-65°C to +150°C

## ELECTRICAL CHARACTERISTICS

Unless otherwise specified; Supplies = +15V, -15V; V<sub>IN</sub> = Logic Input.  
 HI-300-303 : V<sub>IN</sub> - for Logic "1" = 4V, for Logic "0" = 0.8V  
 HI-304-307 : V<sub>IN</sub> - for Logic "1" = 11V, for Logic "0" = 3.5V

PARAMETER	TEMP	-55°C to +125°C			0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG SWITCH CHARACTERISTICS</b>								
Analog Signal Range	Full	-15		+15	-15		+15	V
R <sub>ON</sub> ON Resistance (Note 2)	+25°C		35	50		35	50	Ω
	Full		40	75		40	75	Ω
I <sub>S</sub> OFF OFF Input Leakage Current (Note 3)	+25°C		0.04	1		0.04	5	nA
	Full		1	100		0.2	100	nA
I <sub>D</sub> OFF OFF Output Leakage Current (Note 3)	+25°C		0.04	1		0.04	5	nA
	Full		1	100		0.2	100	nA
I <sub>D</sub> ON ON Leakage Current (Note 4)	+25°C		0.03	1		0.03	5	nA
	Full		0.5	100		0.2	100	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>INL</sub> Input Low Level *	Full			0.8			0.8	V
V <sub>INH</sub> Input High Level *	Full	4			4			V
V <sub>INL</sub> Input Low Level **	Full			3.5			3.5	V
V <sub>INH</sub> Input High Level **	Full	11			11			V
I <sub>INL</sub> Input Leakage Current (Low) (Note 5)	Full			1			1	μA
I <sub>INH</sub> Input Leakage Current (High) (Note 5)	Full			1			1	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>OPEN</sub> Break-Before-Make Delay ***	+25°C		60			60		ns
t <sub>ON</sub> Switch On Time *	+25°C		210	300		210	300	ns
t <sub>OFF</sub> Switch Off Time *	+25°C		160	250		160	250	ns
t <sub>ON</sub> Switch On Time **	+25°C		160	250		160	250	ns
t <sub>OFF</sub> Switch Off Time **	+25°C		100	150		100	150	ns
Off Isolation (Note 6)	+25°C		60			60		dB
Charge Injection (Note 7)	+25°C		3			3		mV
C <sub>S</sub> OFF Input Switch Capacitance	+25°C		16			16		pF
C <sub>D</sub> OFF Output Switch Capacitance	+25°C		14			14		pF
C <sub>D</sub> ON Output Switch Capacitance	+25°C		35			35		pF
C <sub>IN</sub> (High) Digital Input Capacitance	+25°C		5			5		pF
C <sub>IN</sub> (Low) Digital Input Capacitance	+25°C		5			5		pF
<b>POWER REQUIREMENTS</b>								
I <sup>+</sup> Current * (Note 8)	+25°C		0.09	0.5		0.09	0.5	mA
	Full			1			1	mA
I <sup>-</sup> Current * (Note 8)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA
I <sup>+</sup> Current * (Note 9)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA
I <sup>-</sup> Current * (Note 9)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA
I <sup>+</sup> Current ** (Note 10)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA
I <sup>-</sup> Current ** (Note 10)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA
I <sup>+</sup> Current ** (Note 11)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA
I <sup>-</sup> Current ** (Note 11)	+25°C		0.01	10		0.01	100	μA
	Full			100				μA

\* HI-300 thru HI-303 Only; \*\* HI-304 thru HI-307 Only; \*\*\* HI-301, HI-303, HI-305, HI-307 Only



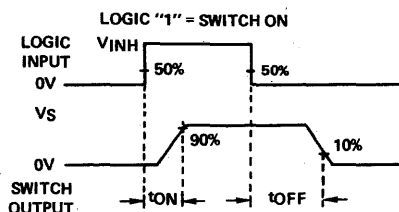
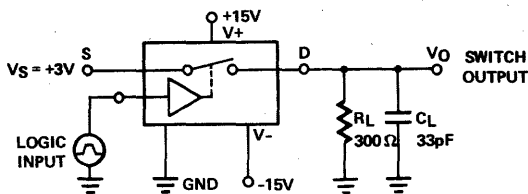
**ELECTRICAL CHARACTERISTICS NOTES:**

1. As with all semiconductors, stresses listed under "Absolute Maximum Ratings" may be applied to devices (one at a time) without resulting in permanent damage. This is a stress rating only. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. The conditions listed under "Electrical Characteristics" are the only conditions recommended for satisfactory operation.
2.  $V_S = \pm 10V$ ,  $I_{OUT} = -10mA$  On resistance derived from the voltage measured across the switch under the above conditions.
3.  $V_S = \pm 14V$ ,  $V_D = \mp 14V$ .
4.  $V_S = V_D = \pm 14V$ .
5. The digital inputs are diode protected MOS gates and typical leakages of 1nA or less can be expected.
6.  $V_S = 1V_{RMS}$ ,  $f = 500kHz$ ,  $C_L = 15pF$ ,  $R_L = 1k$ .  
 $C_L = C_{FIXTURE} + C_{PROBE}$ , "Off Isolation" =  $20 \log V_S / V_D$ .
7.  $V_S = 0V$ ,  $C_L = 10,000pF$ , Logic Drive = 5V pulse. (HI-300-303) Switches are symmetrical; S and D may be interchanged. Logic Drive = 15V (HI-304-307)
8.  $V_{IN} = 4V$  (one input) (all other inputs = 0V)
9.  $V_{IN} = 0.8V$  (all inputs).
10.  $V_{IN} = 15V$  (all inputs).
11.  $V_{IN} = 0V$  (all inputs).
12. To drive from DTL/TTL circuits, pull-up resistors to +5V supply are recommended.

**TEST CIRCUITS**

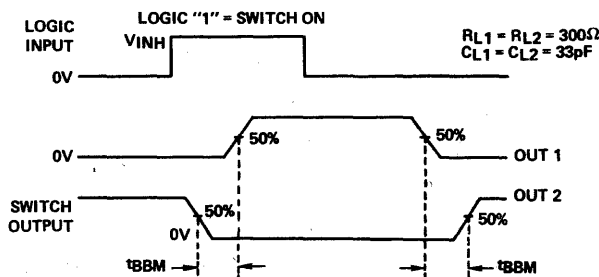
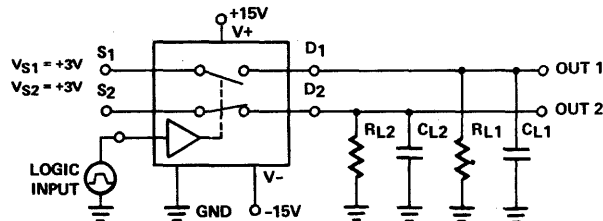
**SWITCHING TEST CIRCUIT ( $t_{ON}$ ,  $t_{OFF}$ )**

SWITCH TYPE	$V_{INH}$
HI-300 thru HI-303	4V
HI-304 thru HI-307	15V



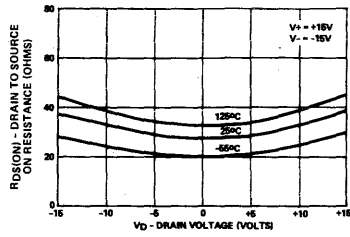
**BREAK-BEFORE-MAKE TEST CIRCUIT ( $t_{BBM}$ )**

SWITCH TYPE	$V_{INH}$
HI-301, HI-303	5V
HI-305, HI-307	15V

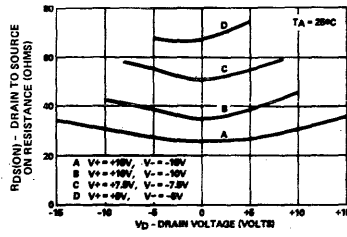




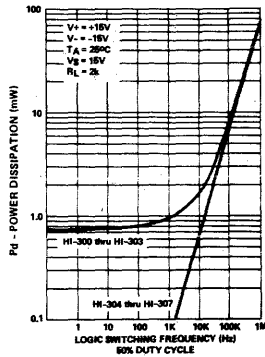
**R<sub>DS(ON)</sub> VS. V<sub>D</sub> AND TEMPERATURE**



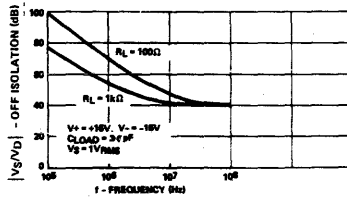
**R<sub>DS(ON)</sub> VS. V<sub>D</sub> AND POWER SUPPLY VOLTAGE**



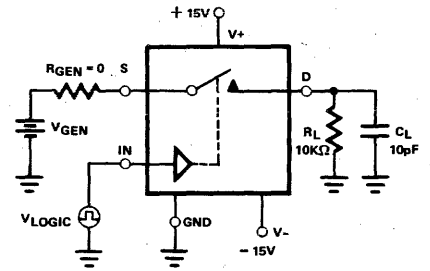
**DEVICE POWER DISSIPATION VS. SWITCHING FREQUENCY SINGLE LOGIC INPUT**



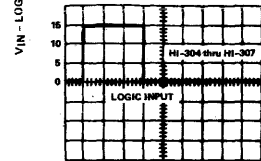
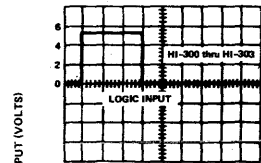
**OFF ISOLATION VS. FREQUENCY**



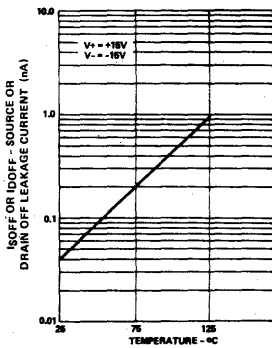
Typical delay, rise, fall, settling times, and switching transients in this circuit.



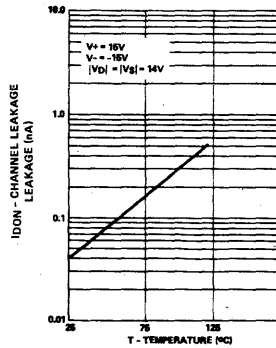
If R<sub>GEN</sub>, R<sub>L</sub> or C<sub>L</sub> is increased, there will be proportional increases in rise and/or fall RC times.



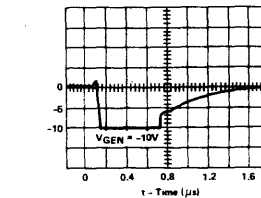
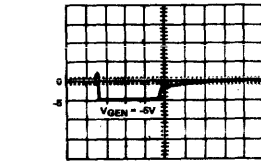
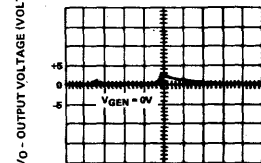
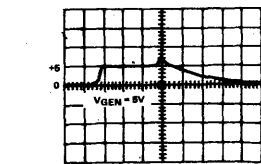
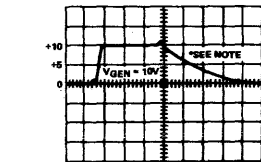
**I<sub>S(OFF)</sub> OR I<sub>D(OFF)</sub> VS. TEMPERATURE \***



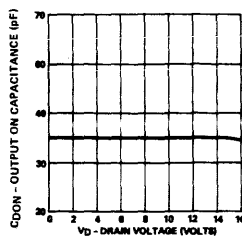
**I<sub>D(ON)</sub> VS. TEMPERATURE \***



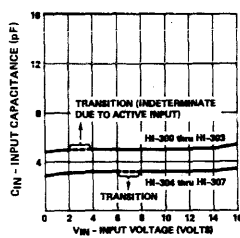
\* The net leakage into the source or drain is the n-channel leakage minus the p-channel leakage. This difference can be positive, negative, or zero depending on the analog voltage and temperature, and will vary greatly from unit to unit.



**OUTPUT ON CAPACITANCE VS. DRAIN VOLTAGE**

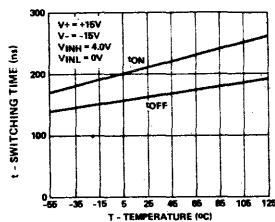


**DIGITAL INPUT CAPACITANCE VS. INPUT VOLTAGE**

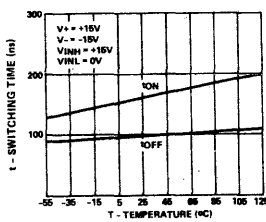




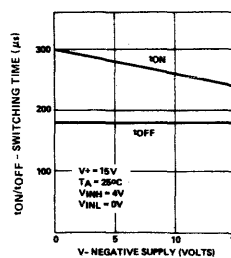
**SWITCHING TIME VS. TEMPERATURE**  
HI-300 thru HI-303



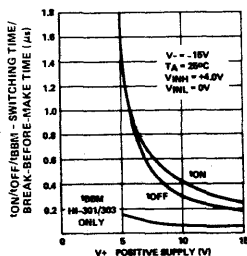
**SWITCHING TIME VS. TEMPERATURE**  
HI-304 thru HI-307



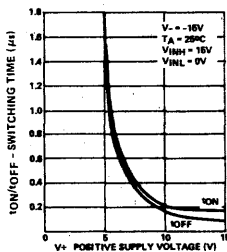
**SWITCHING TIME VS. NEGATIVE SUPPLY VOLTAGE**  
HI-300 thru HI-303



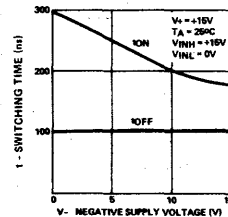
**SWITCHING TIME AND BREAK BEFORE MAKE TIME VS. POSITIVE SUPPLY VOLTAGE**  
HI-300 thru HI-303



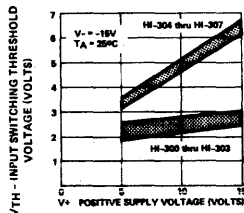
**SWITCHING TIME VS. POSITIVE SUPPLY VOLTAGE**  
HI-304 thru HI-307



**SWITCHING TIME VS. NEGATIVE SUPPLY VOLTAGE**  
HI-304 thru HI-307



**INPUT SWITCHING THRESHOLD VS. POSITIVE SUPPLY VOLTAGE**  
HI-300 thru HI-307



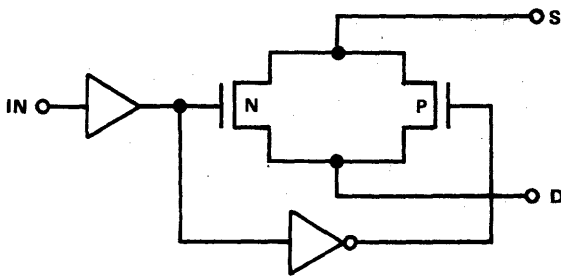


**FEATURES**

- ANALOG SIGNAL RANGE ( $\pm 15V$  SUPPLIES)  $\pm 15V$
- LOW LEAKAGE (TYP. @ 25°C) 40pA
- LOW LEAKAGE (TYP. @ 125°C) 1nA
- LOW ON RESISTANCE (TYP. @ 25°C) 35 $\Omega$
- BREAK-BEFORE-MAKE DELAY (TYP.) 60ns
- CHARGE INJECTION 30pC
- TTL COMPATIBLE
- SYMMETRICAL SWITCH ELEMENTS
- LOW OPERATING POWER (TYP.) 1.0mW

**APPLICATIONS**

- SAMPLE AND HOLD i.e. LOW LEAKAGE SWITCHING
- OP AMP GAIN SWITCHING i.e. LOW ON RESISTANCE
- PORTABLE BATTERY OPERATED CIRCUITS
- LOW LEVEL SWITCHING CIRCUITS
- DUAL OR SINGLE SUPPLY SYSTEMS

**FUNCTIONAL DIAGRAM**


TYPICAL SWITCH - 300 SERIES

**DESCRIPTION**

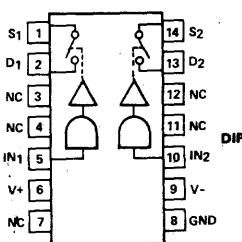
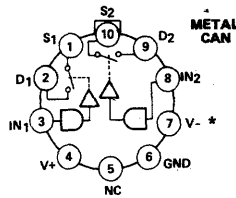
The HI-381 through HI-390 series of switches are monolithic devices fabricated using CMOS technology and the Harris dielectric isolation process. These devices are TTL compatible and are available in four switching configurations. (See device pinout for particular switching function with a logic "1" input.)

These switches feature low leakage and supply currents, low and nearly constant ON resistance over the analog signal range, break-before-make switching and low power dissipation.

The HI-381 and HI-387 switches are available in a 14 pin epoxy or ceramic DIP or 10 pin metal can. The HI-384 and HI-390 are available in a 16 pin epoxy or ceramic DIP. Each of the individual switch types are available in the -55°C to +125°C and 0°C to +75°C operating ranges.

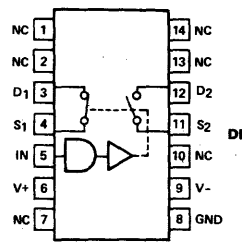
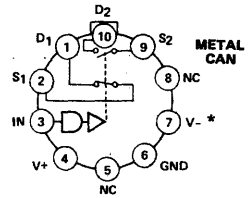
**PINOUTS** (SWITCH STATES ARE FOR A LOGIC "1" INPUT)

Section 11 for Packaging


**DUAL SPST HI-381**  
(TOP VIEWS)


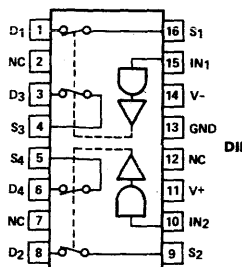
LOGIC	SW 1	SW 2
0	ON	OFF
1	OFF	ON

\*The substrate and case are internally tied to V-. (The case should not be used as the V- connection, however.)

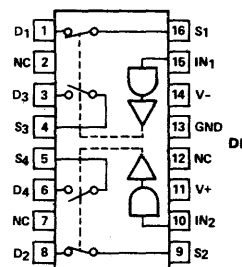

**SPDT HI-387**  
(TOP VIEWS)


LOGIC	SW 1	SW 2
0	OFF	ON
1	ON	OFF

\*The substrate and case are internally tied to V-. (The case should not be used as the V- connection, however.)


**DUAL DPST HI-384**  
(TOP VIEW)

LOGIC	SW 1-4
0	OFF
1	ON


**DUAL SPDT HI-390**  
(TOP VIEW)

LOGIC	SW 1	SW 2	SW 3	SW 4
0	OFF	ON	ON	OFF
1	ON	OFF	OFF	ON

**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Voltage Between Supplies	44V (±22)	Total Power Dissipation	14 Pin Epoxy DIP	526mW
Digital Input Voltage	V <sup>+</sup> +4.0V V <sup>-</sup> -4.0V		14 Pin Ceramic DIP	588mW
Analog Input Voltage	V <sup>+</sup> +1.5V V <sup>-</sup> -1.5V		16 Pin Epoxy DIP	625mW
			16 Pin Ceramic DIP	685mW
			10 Pin Metal Can*	435mW
			*Derate 6.9mW/°C above T <sub>A</sub> = 70°C	
Storage Temperature Range	-65°C to +150°C	Operating Temperature	HI-3XX-2	-55°C to +125°C
			HI-3XX-5	0°C to +75°C

**ELECTRICAL CHARACTERISTICS** Unless otherwise specified; Supplies = +15V, -15V; VIN = Logic Input, VIN for logic "1" = 4V, for logic 0 = .8V

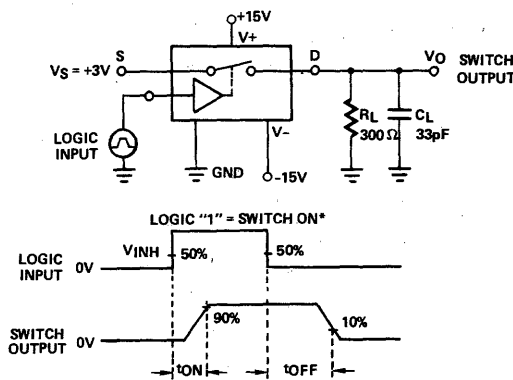
PARAMETER	TEMP	-55°C to +125°C			0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG SWITCH CHARACTERISTICS</b>								
Analog Signal Range	FULL	-15		+15	-15		+15	V
R <sub>ON</sub> ON Resistance (Note 2)	+25°C		35	50		35	50	Ω
	FULL		40	75		40	75	Ω
I <sub>S</sub> OFF Input Leakage Current (Note 3)	+25°C		.04	1		.04	5	nA
	FULL		1	100		0.2	100	nA
I <sub>O</sub> OFF Output Leakage Current (Note 3)	+25°C		.04	1		.04	5	nA
	FULL		1	100		0.2	100	nA
I <sub>D</sub> ON ON Leakage Current (Note 4)	+25°C		.03	1		.03	5	nA
	FULL		0.5	100		0.2	100	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>INL</sub> Input Low Level	FULL			.8			.8	V
V <sub>INH</sub> Input High Level	FULL	4			4			V
I <sub>I</sub> NH Input Leak. Current (High) (Note 5)	FULL			1			1	μA
I <sub>I</sub> NL Input Leak. Current (Low) (Note 5)	FULL			1			1	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>OPEN</sub> , Break-Before Make Delay 390 only)	+25°C		60			60		ns
t <sub>ON</sub> , Switch ON Time	+25°C		210	300		210	300	ns
t <sub>OFF</sub> , Switch OFF Time	+25°C		160	250		160	250	ns
OFF Isolation (Note 6)	+25°C		60			60		dB
Charge Injection (Note 7)	+25°C		3			3		mV
C <sub>S</sub> OFF Input Switch Capacitance	+25°C		16			16		pF
C <sub>D</sub> OFF Output Switch Capacitance	+25°C		14			14		pF
C <sub>D</sub> ON Output Switch Capacitance	+25°C		35			35		pF
C <sub>I</sub> N (High) Digital Input Capacitance	+25°C		5			5		pF
C <sub>I</sub> N (Low) Digital Input Capacitance	+25°C		5			5		pF
<b>POWER REQUIREMENTS</b>								
I <sup>+</sup> Current (Note 8)	+25°C		.09	.5		.09	.5	mA
	FULL			1			1	mA
I <sup>-</sup> Current (Note 8)	+25°C		.01	10		.01	100	μA
	FULL			100			100	μA
I <sup>+</sup> Current (Note 9)	+25°C		.01	10		.01	100	μA
	FULL			100			100	μA
I <sup>-</sup> Current (Note 9)	+25°C		.01	10		.01	100	μA
	FULL			100			100	μA

**ELECTRICAL CHARACTERISTICS NOTES :**

- As with all semiconductors, stresses listed under "Absolute Maximum Ratings" may be applied to devices (one at a time) without resulting in permanent damage. This is a stress rating only. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. The conditions listed under "Electrical Characteristics" are the only conditions recommended for satisfactory operation.
- $V_S = \pm 10V$ ,  $I_{OUT} = -10mA$  on resistance derived from the voltage measured across the switch under the above conditions.
- $V_S = \pm 14V$ ,  $V_D = \mp 14V$ .
- $V_S = V_D = \pm 14V$ .
- The digital inputs are diode protected MOS gates and typical leakages of 1nA or less can be expected.
- $V_S = 1V_{RMS}$ ,  $f = 500kHz$ ,  $C_L = 15pF$ ,  $R_L = 1k$ ,  $C_L = C_{FIXTURE} + C_{PROBE}$ , "off isolation" =  $20 \log V_S/V_D$ .
- $V_S = 0V$ ,  $C_L = 10,000pF$ , Logic Drive = 5V pulse. Switches are symmetrical; S and D may be interchanged.
- $V_{IN} = 4V$ . (one input) (all other inputs = 0)
- $V_{IN} = 0.8V$ . (all inputs)
- To drive from DTL/TTL circuits, pull-up resistors to +5V Supply are recommended.

**TEST CIRCUITS**
**SWITCHING TEST CIRCUIT ( $t_{ON}$ ,  $t_{OFF}$ )**

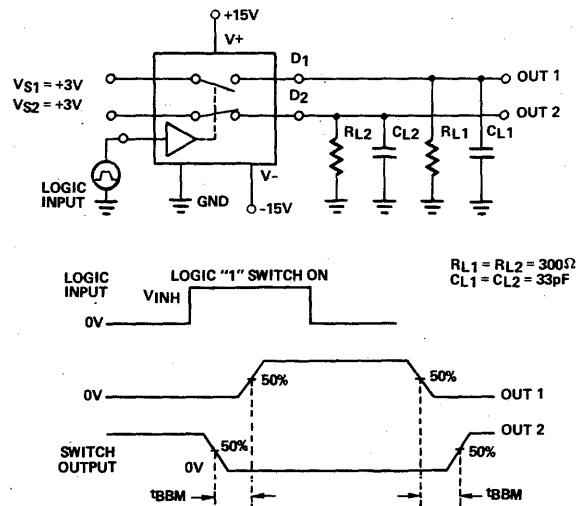
SWITCH TYPE	$V_{INH}$
HI-381 thru HI-390	5V



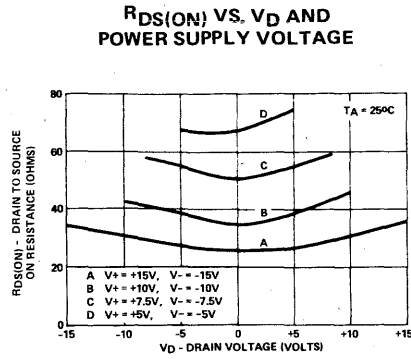
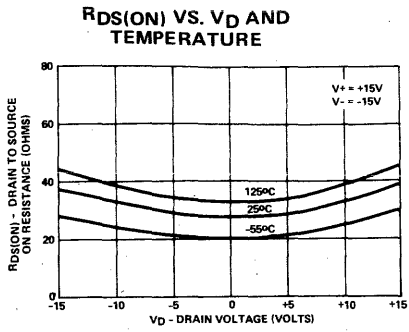
\*Inverted logic for HI-381

**BREAK-BEFORE-MAKE TEST CIRCUIT ( $t_{BBM}$ )**

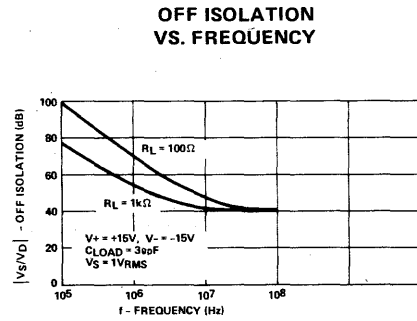
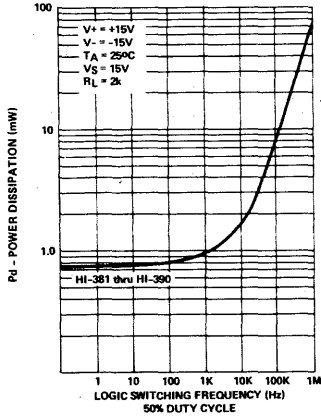
SWITCH TYPE	$V_{INH}$
HI-387 and HI-390	5V



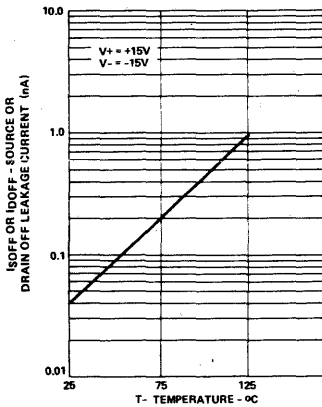
**TYPICAL PERFORMANCE CURVES**



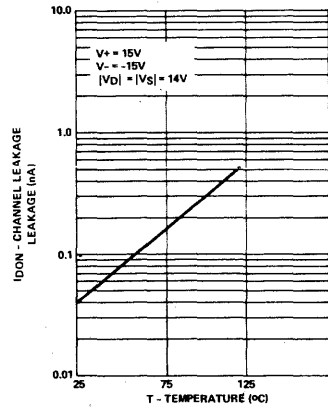
**DEVICE POWER DISSIPATION VS. SWITCHING FREQUENCY SINGLE LOGIC INPUT**



**$I_{SOFF}$  OR  $I_{DOFF}$  VS. TEMPERATURE\***

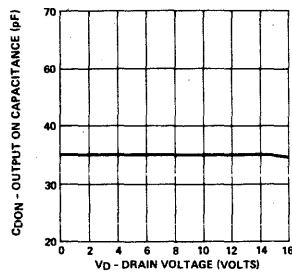


**$I_{DON}$  VS. TEMPERATURE\***

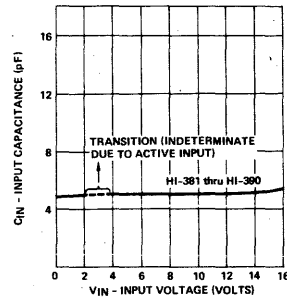


\* The net leakage into the source or drain is the n-channel leakage minus the p-channel leakage. This difference can be positive, negative, or zero depending on the analog voltage and temperature, and will vary greatly from unit to unit.

**OUTPUT ON CAPACITANCE VS. DRAIN VOLTAGE**

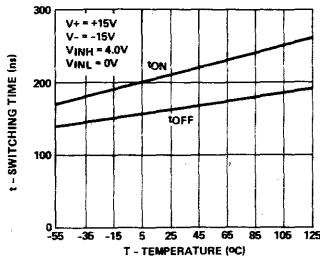


**DIGITAL INPUT CAPACITANCE VS. INPUT VOLTAGE**

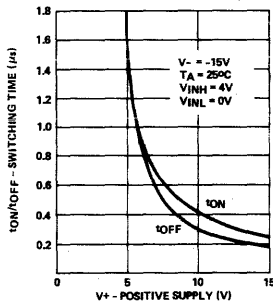




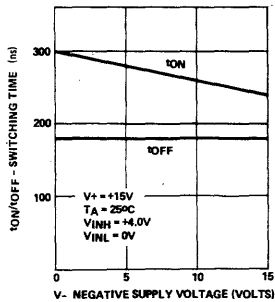
**SWITCHING TIME VS. TEMPERATURE**  
HI-381 thru HI-390



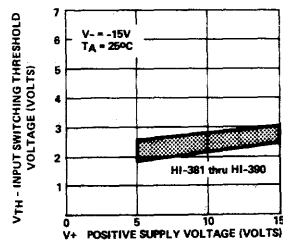
**SWITCHING TIME VS. POSITIVE SUPPLY VOLTAGE**  
HI-381 thru HI-390



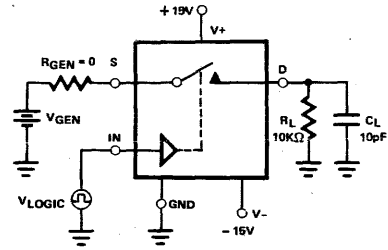
**SWITCHING TIME VS. NEGATIVE SUPPLY VOLTAGE**  
HI-381 thru HI-390



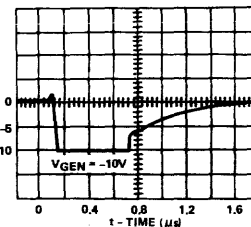
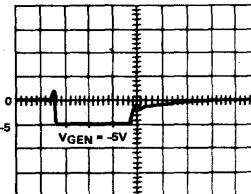
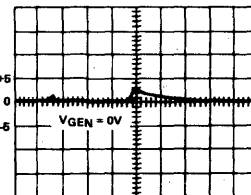
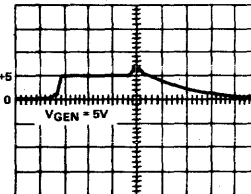
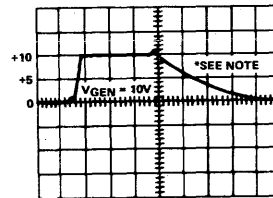
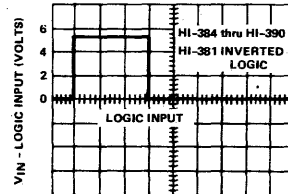
**INPUT SWITCHING THRESHOLD VS. POSITIVE SUPPLY VOLTAGE**  
HI-381 thru HI-390



Typical delay, rise, fall, setting times, and switching transients in this circuit.



If  $R_{GEN}$ ,  $R_L$  or  $C_L$  is increased, there will be proportional increases in rise and/or fall RC times.



\* NOTE: The turn-off time is primarily limited here by the RC time constant (100ns) of the load.



**HARRIS**

# HI-5040 thru HI-5051 HI-5046A and HI-5047A

## CMOS Analog Switches

### FEATURES

- WIDE ANALOG SIGNAL RANGE ±15V
- LOW "ON" RESISTANCE (TYP) 25Ω
- HIGH CURRENT CAPABILITY (TYP) 80mA
- BREAK-BEFORE-MAKE SWITCHING
- TURN-ON TIME (TYP) 370ns
- TURN-OFF TIME (TYP) 280ns
- NO LATCH-UP
- INPUT MOS GATES ARE PROTECTED FROM ELECTROSTATIC DISCHARGE
- DTL, TTL, CMOS, PMOS COMPATIBLE

### APPLICATIONS

- HIGH FREQUENCY SWITCHING
- SAMPLE AND HOLD
- DIGITAL FILTERS
- OP AMP GAIN SWITCHING

### DESCRIPTION

This family of CMOS analog switches offers low-resistance switching performance for analog voltages up to the supply rails and for signal currents up to 80mA. "ON" resistance is low and stays reasonably constant over the full range of operating signal voltage and current.  $R_{ON}$  remains exceptionally constant for input voltages between +5V and -5V and currents up to 50mA. Switch impedance also changes very little over temperature, particularly between 0°C and +75°C.  $R_{ON}$  is nominally 25 ohms for HI-5048 through HI-5051 and HI-5046A/5047A and 50Ω for HI-5040 through HI-5047.

All devices provide break-before-make switching and are TTL and CMOS compatible for maximum application versatility. Performance is further enhanced by Dielectric Isolation processing which insures latch-free operation with very low input and output leakage currents (0.8nA at 25°C). This family of switches also features very low power operation (1.5mW at 25°C).

There are 14 devices in this switch series which are differentiated by type of switch action and value of  $R_{ON}$  (see Functional diagram). All devices are available in 16 pin D.I.P. packages. The HI-5040/5050 switches can directly replace IH-5040 series devices and are functionally compatible with the DG 180/190 family. Each switch type is available in the -55°C to +125°C and 0°C to +75°C performance grades.

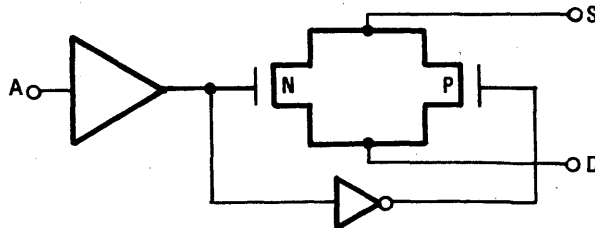
### FUNCTIONAL DESCRIPTION

Section 11 for Packaging

PART NUMBER	TYPE	$R_{ON}$
HI-5040	SPST	75Ω
HI-5041	DUAL SPST	75Ω
HI-5042	SPDT	75Ω
HI-5043	DUAL SPDT	75Ω
HI-5044	DPST	75Ω
HI-5045	DUAL DPST	75Ω
HI-5046	DPDT	75Ω
HI-5046A	DPDT	30Ω
HI-5047	4PST	75Ω
HI-5047A	4PST	30Ω
HI-5048	DUAL SPST	30Ω
HI-5049	DUAL DPST	30Ω
HI-5050	SPDT	30Ω
HI-5051	DUAL SPDT	30Ω

### FUNCTIONAL DIAGRAM

TYPICAL DIAGRAM



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage ( $V^+ - V^-$ )	36V	Analog Current (S to D)	80mA
$V_R$ to Ground	$V^+, V^-$	Total Power Dissipation*	450mW
Digital and Analog	$V^+ +4V$	Operating Temperature	
Input Voltage	$V^- -4V$	HI-50XX-2	-55°C to +125°C
		HI-50XX-5	0°C to +75°C
		Storage Temperature	-65°C to +150°C

\*Derate 6mW/°C above  $T_A = 75^\circ\text{C}$

## ELECTRICAL CHARACTERISTICS

Unless Otherwise Specified

Supplies = +15V, -15V;  $V_R = 0V$ ;  $V_{AH}$  (Logic Level High) = 3.0V;  $V_{AL}$  (Logic Level Low) = +0.8V,  $V_L = +5V$

For Test Conditions, consult Performance Characteristics

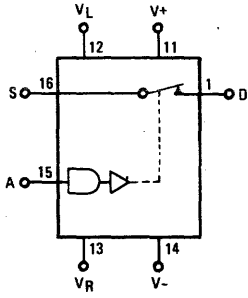
PARAMETER	TEMP	-55°C to +125°C			0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG SWITCH CHARACTERISTICS</b>								
Analog Signal Range	Full	-15		+15	-15		+15	V
$R_{on}$ , "ON" Resistance (Note 1a)	+25°C		50			50		$\Omega$
	Full			75			75	$\Omega$
$R_{on}$ , "ON" Resistance (Note 1b)	+25°C		25			25		$\Omega$
	Full			50			50	$\Omega$
$R_{on}$ , Channel-to-Channel Match (Note 1a)	+25°C		2	10		2	10	$\Omega$
$R_{on}$ , Channel-to-Channel Match (Note 1b)	+25°C		1	5		1	5	$\Omega$
$I_S$ (OFF) = $I_D$ (OFF), Off Input or Output Leakage Current	+25°C		0.8			0.8		nA
	Full		100	500		100	500	nA
$I_D$ (ON), On Leakage Current	+25°C		0.01			0.01		nA
	Full		2	500		2	500	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
$V_{AL}$ , Input Low Threshold	Full			0.8			0.8	V
$V_{AH}$ , Input High Threshold	Full	3.0			3.0			V
$I_A$ , Input Leakage Current (High or Low)	Full		.01	1.0		.01	1.0	$\mu\text{A}$
<b>SWITCHING CHARACTERISTICS</b>								
$t_{on}$ , Switch "ON" Time	+25°C		370	1000		370	1000	ns
$t_{off}$ , Switch "OFF" Time	+25°C		280	500		280	500	ns
Charge Injection (Note 2)	+25°C		5	20		5		mV
"OFF Isolation" (Note 3)	+25°C	75	80			80		dB
"Crosstalk" (Note 3)	+25°C	80	88			88		dB
$C_S$ (OFF), Input Switch Capacitance	+25°C		11			11		pF
$C_D$ (OFF), Output Switch Capacitance	+25°C		11			11		pF
$C_D$ (ON), Output Switch Capacitance	+25°C		22			22		pF
$C_A$ , Digital Input Capacitance	+25°C		5			5		pF
$C_{DS}$ (OFF), Drain-To-Source Capacitance	+25°C		0.5			0.5		pF
<b>POWER REQUIREMENTS</b>								
$P_D$ , Quiescent Power Dissipation	+25°C		1.5			1.5		mW
$I^+$ , +15V Quiescent Current	Full			0.3			0.5	mA
$I^-$ , -15V Quiescent Current	Full			0.3			0.5	mA
$I_L$ , +5V Quiescent Current	Full			0.3			0.5	mA
$I_R$ , Gnd Quiescent Current	Full			0.3			0.5	mA

- NOTES: 1.  $V_{OUT} = \pm 10V$ ,  $I_{OUT} = 1mA$   
a) For HI-5040 thru HI-5047  
b) For HI-5048 thru HI-5051, HI-5046A/5047A  
2.  $V_{IN} = 0V$ ,  $C_L = 10,000pF$   
3.  $R_L = 100\Omega$ ,  $f = 100\text{KHz}$ ,  $V_{IN} = 2V_{PP}$ ,  $C_L = 5pF$

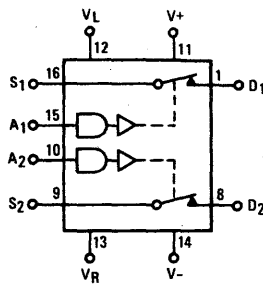
**SWITCH FUNCTIONS**

SWITCH STATES ARE FOR LOGIC "1" INPUT

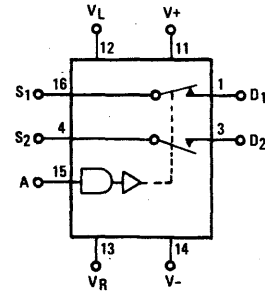
**SPST**  
HI-5040 (75Ω)



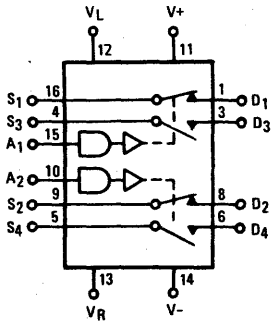
**DUAL SPST**  
HI-5041 (75Ω)



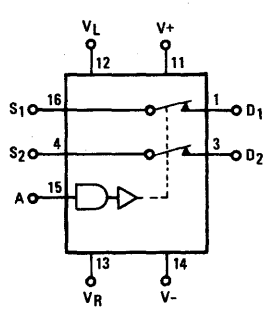
**SPDT**  
HI-5042 (75Ω)



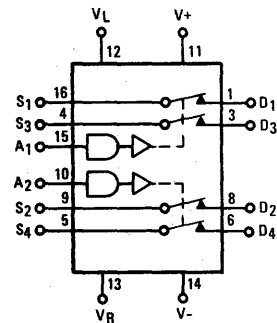
**DUAL SPDT**  
HI-5043 (75Ω)



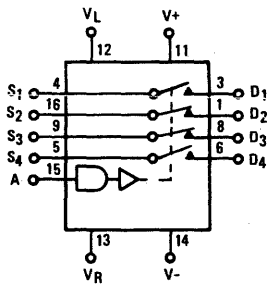
**DPST**  
HI-5044 (75Ω)



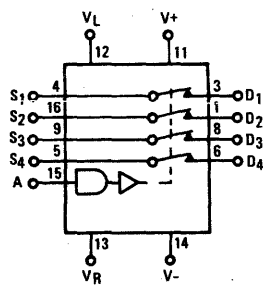
**DUAL DPST**  
HI-5045 (75Ω)



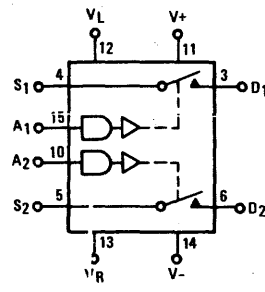
**DPDT**  
HI-5046 (75Ω)  
HI-5046A (30Ω)



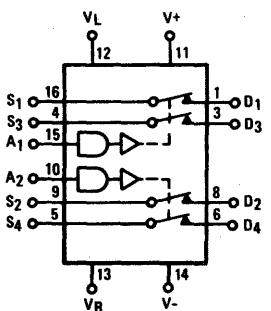
**4PST**  
HI-5047 (75Ω)  
HI-5047A (30Ω)



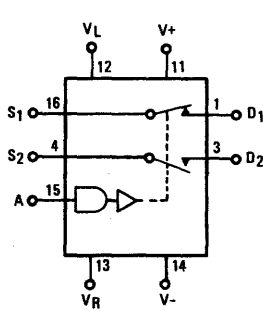
**DUAL SPST**  
HI-5048 (30Ω)



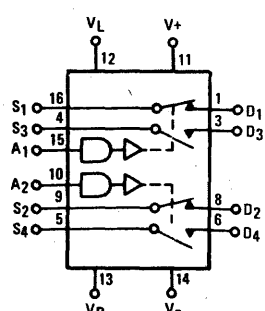
**DUAL DPST**  
HI-5049 (30Ω)



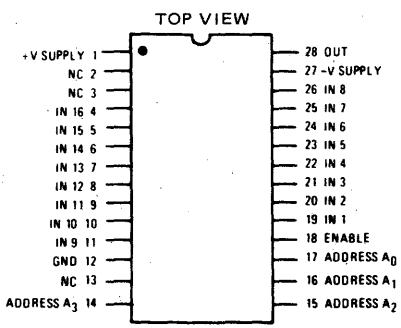
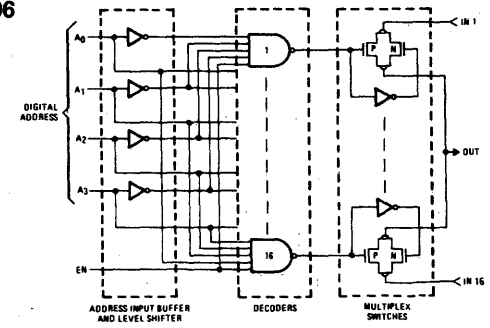
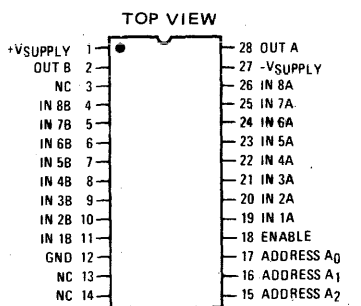
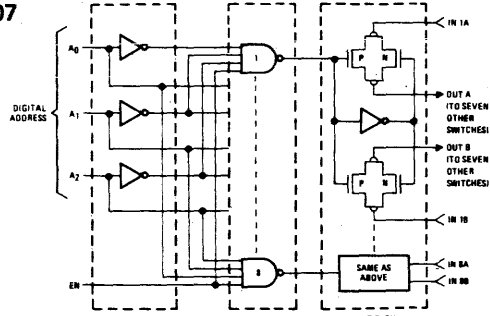
**SPDT**  
HI-5050 (30Ω)



**DUAL SPDT**  
HI-5051 (30Ω)





FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• LOW ON RESISTANCE (TYP.) 170Ω</li> <li>• WIDE ANALOG SIGNAL RANGE ±15V</li> <li>• DIRECTLY TTL/CMOS COMPATIBLE 2.4V (LOGIC "1")</li> <li>• ACCESS TIME (TYP.) 300ns</li> <li>• HIGH CURRENT CAPABILITY (TYP.) 50mA</li> <li>• BREAK-BEFORE-MAKE SWITCHING</li> <li>• NO LATCH-UP</li> </ul>	<p>These monolithic CMOS multiplexers each include an array of eight analog switches, a digital decode circuit for channel selection, a voltage reference for logic thresholds, and an ENABLE input for device selection when several multiplexers are present.</p> <p>The Dielectric Isolation (DI) process used in fabrication of these devices eliminates the problem of latchup. Also, DI offers much lower substrate leakage and parasitic capacitance than conventional junction-isolated CMOS (See Application Note 521). With the low ON resistance (180Ω typical), this allows low static error, fast channel switching rates, and fast settling.</p> <p>Switches are guaranteed to break-before-make, so two channels are never shorted together.</p> <p>The switching threshold for each digital input is established by an internal +5V reference, providing a guaranteed minimum 2.4V for "1" and maximum 0.8V for "0". This allows direct interface without pullup resistors to signals from most logic families: CMOS, TTL, DTL and some PMOS. For protection against transient overvoltage, the digital inputs include a series 200Ω resistor and a diode clamp to each supply.</p> <p>The HI-506 is a sixteen channel single-ended multiplexer, and the HI-507 is an eight channel differential version. Each device is packaged in a 28 pin DIP. The recommended supply voltage is ±15V, and reasonable performance is available down to ±7V; however, a power-up reset routine may be required for operation below ±10V. For more information, request Harris Analog Tech Brief # 20.</p> <p>The HI-506/507 are specified for operation from 0°C to +70°C. The "-2" versions are specified from -55°C to +125°C. "Dash 8" (-8) designates -2 parts which have been screened per Mil-Std-883/Method 5004/Class B.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• DATA ACQUISITION SYSTEMS</li> <li>• PRECISION INSTRUMENTATION</li> <li>• DEMULTIPLEXING</li> <li>• SELECTOR SWITCH</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p><b>HI-506</b></p> <p>TOP VIEW</p> 	<p><b>HI-506</b></p> 
<p><b>HI-507</b></p> <p>TOP VIEW</p> 	<p><b>HI-507</b></p> 



# SPECIFICATIONS

INTERFACE

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage Between Pins 1 and 27 40V  
 V<sub>EN</sub>, V<sub>A</sub>, Digital Input Overvoltage:  
 V<sub>A</sub> { V<sub>Supply</sub> (+) +4V  
       V<sub>Supply</sub> (-) -4V  
 Analog Input Overvoltage: (Note 6)  
 V<sub>D</sub> or V<sub>S</sub> { V<sub>Supply</sub> (+) +2V  
               V<sub>Supply</sub> (-) -2V

Total Power Dissipation\* 1200 mW  
 Operating Temperature:  
 HI-506/HI-507-2 -55°C to +125°C  
 HI-506/HI-507-5 0°C to +75°C  
 Storage Temperature -65°C to +150°C

\*Derate 19.7mW/°C above T<sub>A</sub> = 110°C

## ELECTRICAL CHARACTERISTICS Unless Otherwise Specified: Supplies = +15V, -15V; V<sub>AH</sub>(Logic Level High) = +2.4V, V<sub>AL</sub>(Logic Level Low) = +0.8V. For Test Conditions, consult Performance Characteristics section.

PARAMETER	TEMP	HI-506/HI-507-2 -55°C to +125°C			HI-506/HI-507-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
* V <sub>S</sub> , Analog Signal Range	Full	-15		+15	-15		+15	V
* R <sub>ON</sub> , On Resistance (Note 1)	+25°C Full		170 *	300 400		270 *	400 500	Ω Ω
* ΔR <sub>ON</sub> , (Between Channels)	+25°C		6			6		%
* I <sub>S(OFF)</sub> , Off Input Leakage Current	+25°C Full		0.03	±50		0.03	±50	nA nA
* I <sub>D(OFF)</sub> , Off Output Leakage Current HI-506 HI-507	+25°C Full Full		0.3	±500 ±250		1.0	±500 ±250	nA nA nA
* I <sub>D(ON)</sub> , On Channel Leakage Current HI-506 HI-507	+25°C Full Full		0.3	±500 ±250		1.0	±500 ±250	nA nA nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>AL</sub> , Input Low Threshold	Full			+0.8			+0.8	V
V <sub>AH</sub> , Input High Threshold	Full	+2.4			+2.4			V
* I <sub>A</sub> , Input Leakage Current (High or Low)(Note 2)	Full			1.0			5.0	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>A</sub> , Access Time	+25°C		300	1000		300		ns
t <sub>OPEN</sub> , Break-Before Make Delay	+25°C		80			80		ns
t <sub>ON(EN)</sub> , Enable Delay (ON)	+25°C		300	1000		300		ns
t <sub>OFF(EN)</sub> , Enable Delay (OFF)	+25°C		300	1000		300		ns
Settling Time (0.1%) (0.0 25%)	+25°C +25°C		1.2 2.4			1.2 2.4		μs μs
"Off Isolation" (Note 3)	+25°C		75			75		dB
C <sub>S(OFF)</sub> , Channel Input Capacitance	+25°C		4			4		pF
C <sub>D(OFF)</sub> , Channel Output Capacitance HI-506 HI-507	+25°C +25°C		44 22			44 22		pF pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		2.2			2.2		pF
C <sub>DS(OFF)</sub> , Input to Output Capacitance	+25°C		0.08			0.08		pF
<b>POWER REQUIREMENTS</b>								
* I <sub>+</sub> , Current Pin 1 (Note 4)	Full		1.7	3.0		3.4	5.0	mA
* I <sub>-</sub> , Current Pin 27 (Note 4)	Full		0.4	1.0		0.8	2.0	mA
* I <sub>+</sub> , Standby (Note 5)	Full		1.7	3.0		3.4	5.0	mA
* I <sub>-</sub> , Standby (Note 5)	Full		0.4	1.0		0.8	2.0	mA

- NOTES: 1. V<sub>OUT</sub> = ±10V, I<sub>OUT</sub> = -1mA  
 2. Digital Inputs are Mos Gates. Typical Leakage Less Than 1nA.  
 3. V<sub>EN</sub> = 0.8V, R<sub>L</sub> = 1K, C<sub>L</sub> = 28pF, V<sub>S</sub> = 7VRMS, f = 500kHz.  
 4. V<sub>EN</sub> = 4.0V, All V<sub>A</sub> = 4.0V  
 5. V<sub>EN</sub> = 0V, All V<sub>A</sub> = 0V  
 6. If Analog Input Overvoltage Conditions are Anticipated, Use of HI-506A/507A Protected Multiplexers is Recommended. See HI-506A/507A Data Sheet.

\* 100% Tested for Dash 8 at +25°C and +125°C Only.

## TRUTH TABLES

### HI-506

A <sub>3</sub>	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	"ON" CHANNEL
X	X	X	X	L	NONE
L	L	L	L	H	1
L	L	L	L	H	2
L	L	L	H	H	3
L	L	H	L	H	4
L	H	L	L	H	5
L	H	L	H	H	6
L	H	H	L	H	7
L	H	H	H	H	8
H	L	L	L	H	9
H	L	L	H	H	10
H	L	H	L	H	11
H	L	H	H	H	12
H	H	L	L	H	13
H	H	L	H	H	14
H	H	H	L	H	15
H	H	H	H	H	16

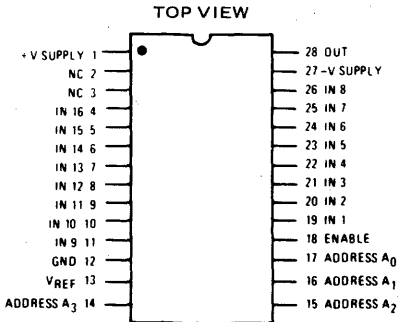
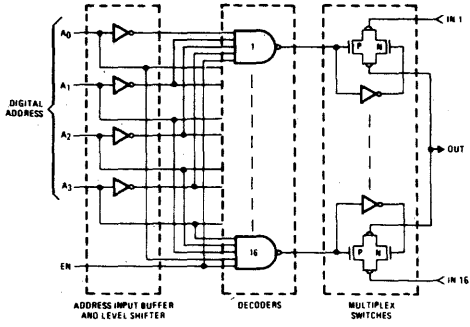
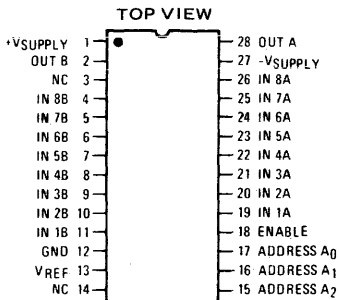
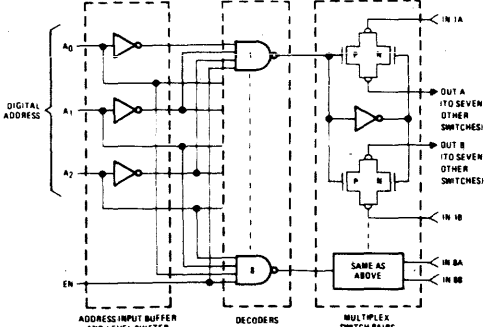
### HI-507

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH PAIR
X	X	X	L	NONE
L	L	L	H	1
L	L	L	H	2
L	L	H	H	3
L	H	L	H	4
H	L	L	H	5
H	L	H	H	6
H	H	L	H	7
H	H	H	H	8

Harris Semiconductor

# HI-506A/HI-507A

## 16 Channel CMOS Analog Multiplexer with Overvoltage Protection

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• ANALOG/DIGITAL OVERVOLTAGE PROTECTION</li> <li>• FAIL SAFE WITH POWER LOSS (NO LATCHUP)</li> <li>• BREAK-BEFORE-MAKE SWITCHING</li> <li>• DTL/TTL AND CMOS COMPATIBLE</li> <li>• ANALOG SIGNAL RANGE <math>\pm 15V</math></li> <li>• ACCESS TIME (TYP.) 500ns</li> <li>• SUPPLY CURRENT AT 1MHz ADDRESS TOGGLE (TYP.) 4mA</li> <li>• STANDBY POWER (TYP.) 7.5mW</li> </ul>	<p>The HI-506A and HI-507A are dielectrically isolated CMOS analog multiplexers incorporating an important feature; they withstand analog input voltages much greater than the supplies. This is essential in any system where the analog inputs originate outside the equipment. They can withstand a continuous input up to 10 volts greater than either supply, which eliminates the possibility of damage when supplies are off, but input signals are present. Equally important they can withstand brief input transient spikes of several hundred volts; which otherwise would require complex external protection networks. Necessarily, ON resistance is somewhat higher than similar unprotected devices, but very low leakage currents combine to produce low errors. Application Notes 520 and 521 further explain these features.</p> <p>The HI-506A-2 and HI-507A-2 are specified over <math>-55^{\circ}C</math> to <math>+125^{\circ}C</math> while the -5 versions are specified over <math>0^{\circ}C</math> to <math>+75^{\circ}C</math>.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• DATA ACQUISITION</li> <li>• INDUSTRIAL CONTROLS</li> <li>• TELEMETRY</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p><b>HI-506A</b></p> 	<p><b>HI-506A</b></p> 
<p><b>HI-507A</b></p> 	<p><b>HI-507A</b></p> 



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage Between Pins 1 and 27	40V	Total Power Dissipation	1200mW
VREF to Ground V+ to Ground	+20V	Operating Temperature	
VEN, VA, Digital Input Overvoltage:		HI-506A/507A-2	-55°C to +125°C
VA { VSupply (+)	+4V	HI-506A/507A-5	0°C to +75°C
VSupply (-)	-4V	Storage Temperature	-65°C to +150°C
Analog Overvoltage:			
VS { VSupply (+)	+20V		
VSupply (-)	-20V		

**ELECTRICAL CHARACTERISTICS (Unless otherwise specified)**

Supplies = +15V, -15V; VREF (Pin 13) = Open; VAH (Logic Level High) = +4.0V; VAL (Logic Level Low) = +0.8V  
For Test Conditions, consult Performance Characteristics section.

PARAMETER	TEMP.	HI-506A/507A-2 -55°C to +125°C			HI-506A/507A-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
*VS, Analog Signal Range	Full	-15		+15	-15		+15	V
*RON, On Resistance (Note 1)	+25°C		1.2	1.5		1.5	1.8	KΩ
	Full		1.5	2.0		1.8	2.0	KΩ
*IS (OFF), Off Input Leakage Current	+25°C		0.03			0.03		nA
	Full			±50			±50	nA
*ID (OFF), Off Output Leakage Current	+25°C		1.0			1.0		nA
	Full			±500			±500	nA
	Full			±250			±250	nA
*ID (OFF) with Input Overvoltage Applied (Note 2)	+25°C		4.0			4.0		nA
	Full			2.0				μA
*ID (ON), On Channel Leakage Current	+25°C		0.1			0.1		nA
	Full			±500			±500	nA
	Full			±250			±250	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
VAL, Input Low Threshold   TTL Drive	Full		0.8			0.8		V
VAH, Input High Threshold   (Note 7)	Full	4.0			4.0			V
VAL   MOS Drive (Note 3)	+25°C		0.8			0.8		V
VAH	+25°C	6.0			6.0			V
*IA, Input Leakage Current (High or Low)	Full		1.0			5.0		μA
<b>SWITCHING CHARACTERISTICS</b>								
tA, Access Time	+25°C		0.5	1.0		0.5		μs
tOPEN, Break-Before Make Delay	+25°C		80			80		ns
tON (EN), Enable Delay (ON)	+25°C		300			300		ns
tOFF (EN), Enable Delay (OFF)	+25°C		300			300		ns
Settling Time (0.1%)	+25°C		1.3			1.3		μs
(0.025%)	+25°C		4.4			4.4		μs
"Off Isolation" (Note 4)	+25°C		65			65		dB
CS (OFF), Channel Input Capacitance	+25°C		5			5		pF
CD (OFF), Channel Output Capacitance	+25°C		50			50		pF
	+25°C		25			25		pF
CA, Digital Input Capacitance	+25°C		5			5		pF
CDS (OFF), Input to Output Capacitance	+25°C		0.1			0.1		pF
<b>POWER REQUIREMENTS</b>								
PD, Power Dissipation	Full		7.5			7.5		mW
*I+, Current Pin 1 (Note 5)	Full		0.5	2.0		0.5	5.0	mA
*I-, Current Pin 27 (Note 5)	Full		0.02	1.0		0.02	2.0	mA
*I+, Standby (Note 6)	Full		0.5	2.0		0.5	5.0	mA
*I-, Standby (Note 6)	Full		0.02	1.0		0.02	2.0	mA

NOTES: 1. VOUT = +10V, IOUT = -100 μA. 2. Analog Overvoltage = ±33V. 3. VREF = +10V. 4. VEN = 0.8V, RL = 1K, CL = 7pF, VS = 3VRMS, f = 500KHz. 5. VEN = +4.0V. 6. VEN = 0.8V. 7. To drive from DTL/TTL circuits, 1KΩ pull-up resistors to +5.0V supply are recommended. \* 100% Tested for Dash 8 at +25°C and +125°C Only.

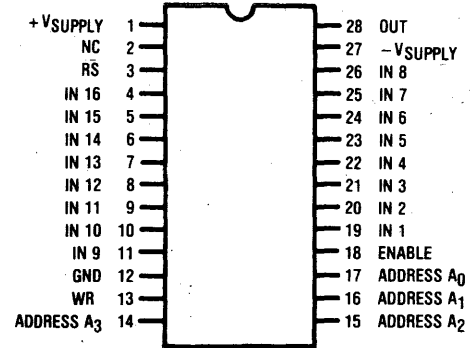
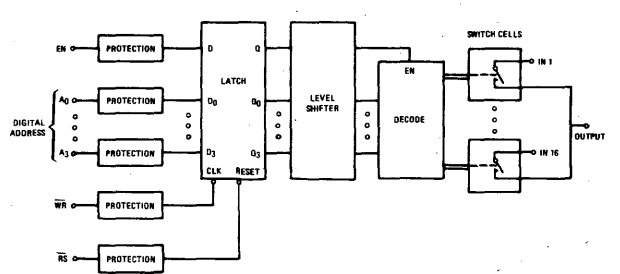
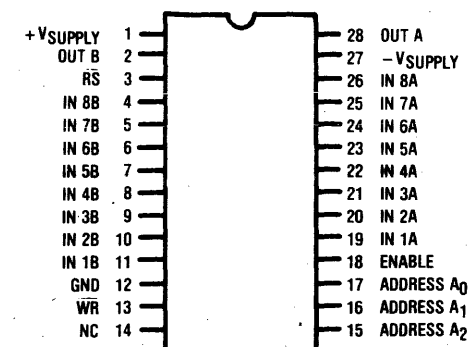
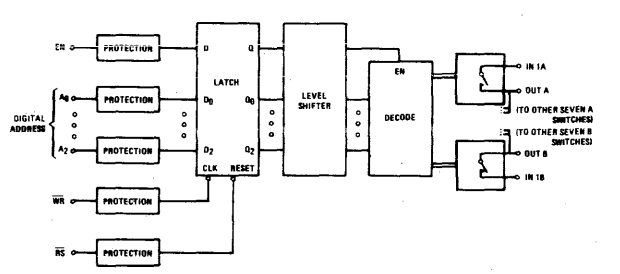
**TRUTH TABLES**

**HI-506A**

A3	A2	A1	A0	EN	"ON" CHANNEL
X	X	X	X	L	NONE
L	L	L	L	H	1
L	L	L	H	H	2
L	L	H	L	H	3
L	L	H	H	H	4
L	H	L	L	H	5
L	H	L	H	H	6
L	H	H	L	H	7
L	H	H	H	H	8
H	L	L	L	H	9
H	L	L	L	H	10
H	L	L	H	H	11
H	L	H	L	H	12
H	L	H	H	H	13
H	H	L	L	H	14
H	H	L	H	H	15
H	H	H	L	H	16

**HI-507A**

A2	A1	A0	EN	ON SWITCH PAIR
X	X	X	L	NONE
L	L	L	H	1
L	L	H	H	2
L	H	L	H	3
L	H	H	H	4
H	L	L	H	5
H	L	H	H	6
H	H	L	H	7
H	H	H	H	8

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• Analog Overvoltage protection</li> <li>• Resettable Latches (<math>\overline{RS}</math>)</li> <li>• TTL/DTL and CMOS Compatible</li> <li>• Failsafe for conditions of Overvoltage &amp; Loss of Power</li> <li>• No SCR Latch-up</li> <li>• Break-before-make switching</li> <li>• Microprocessor Bus compatible</li> <li>• Very low leakage - <math>I_{D(off)} \leq 8nA</math> (typ) over full temp range</li> <li>• Access time - <math>t_A = 500nS</math> (typ)</li> <li>• Minimum write pulse width (<math>\overline{WR}</math>) = 300 nS</li> <li>• OFF isolation = -100dB, typ @ 10kHz</li> </ul>	<p>These monolithic CMOS multiplexers feature on-board address latches, plus overvoltage protection for the analog inputs and the output as well. Each mode accommodates digital inputs for channel selection and an Enable input for device selection under program control. In addition, Write (<math>\overline{WR}</math>) and Reset (<math>\overline{RS}</math>) inputs allow the program to store or clear the channel address.</p> <p>The overvoltage performance of these multiplexers is particularly useful in redundant systems, where the inputs and output should present a very high impedance when power is off. This is achieved by a switch cell with three MOSFET's in series, rather than the conventional single transmission gate design.</p> <p>Each channel can withstand a DC overvoltage up to a magnitude of 10VDC greater than the power supply (or to <math>\pm 25VDC</math> with power off). An OFF channel remains OFF for this condition. If the channel is ON, the output voltage is clamped below the supply rail, which protects the load circuit.</p> <p>The HI-506L offers 16 single-ended channels, and the HI-507L is an 8 channel differential version. The recommended supply voltages are <math>\pm 15V</math>, through operation at reduced levels or with a single supply may also be implemented. The package is a 28 pin ceramic or plastic DIP.</p> <p>Each product is specified for the commercial temperature range (0°C to +75°C, -5 suffix) and the military range (-55°C to +125°C, -2 suffix). Military parts screened per MIL-STD-883 method 5004, Class B are offered with a "-8" suffix.</p>
PINOUT	FUNCTIONAL DIAGRAM
<p><b>HI-506L</b></p> 	<p><b>HI-506L</b></p> 
<p><b>HI-507L</b></p> 	<p><b>HI-507L</b></p> 

CAUTION: Electronic devices are sensitive to electrostatic discharge. Proper I.C. handling procedures should be followed.

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**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage Between Pins 1 and 27	44V	Operating Temperature	
Digital Input Overvoltage, $V_A$ , $V_{EN}$ , $\overline{V_{RS}}$ , $\overline{V_{WR}}$ :			
V supply (+)	+4V	HI-506L/507L-2	-55°C to 125°C
V supply (-)	-4V	HI-506L/507L-5	0°C to 75°C
Analog Overvoltage		Storage Temperature	-65°C to +150°C
Input to Ground	±25VDC		
Total Power Dissipation* (Package)	1200mW	*Derate-8mW/°C above $T_A = +75°C$	

**ELECTRICAL CHARACTERISTICS** (Unless otherwise specified)

+V supply = 15V, -V supply = -15V,  $V_{AH}$  (Logic High) = 2.0V,  $V_{AL}$  (Logic Low) = 0.8V

PARAMETER	HI-506L/507L-2 -55°C to +125°C				HI-506L/507L-5 0°C to +75°C			UNITS
	TEMP	MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
$V_S$ Analog Signal Range	Full		±10			±10		V
$R_{ON}$ , ON Resistance (Note 1)	+25°C			1.2			1.5	KΩ
	Full			1.8			1.8	KΩ
$\Delta R_{ON}$ , Change in $R_{ON}$ (Note 2) between channels	+25°C		5			5		%
$I_{S(off)}$ , OFF input leakage current	+25°C			10			10	nA
	Full			7		7	50	nA
$I_{D(off)}$ , OFF output leakage current	+25°C			10			10	nA
	Full			8		8	200	nA
	HI-506L			200			200	nA
	HI-507L			100			100	nA
$I_{D(on)}$ , ON Channel leakage current	+25°C		5	10		5	10	nA
	Full			10		10	200	nA
	HI-506L			200			200	nA
	HI-507L			100			100	nA
<b>FAULT CHARACTERISTICS</b>								
$I_{S(off)}$ , with Power OFF	Full		10	1000		10	5000	nA
$I_{S(off)}$ , overvoltage (Note 3)	Full		10	750		10	2500	nA
$I_{D(off)}$ , with input over- voltage applied (Note 3)	+25°C		5			5		nA
	Full			10		10	2500	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
$V_{AL}$ , Input Low Threshold	Full		1.4	0.8		1.4	0.8	V
$V_{AH}$ , Input High Threshold	Full	2.0	1.4		2.0	1.4		V
$I_{AH}$ , Input High Current	Full		10	40		10	40	μA
$I_{AL}$ , Input Low Current	Full		40	200		40	200	μA
<b>DYNAMIC SWITCHING CHARACTERISTICS</b>								
$t_a$ , Access Time	+25°C		0.5	1.0		0.5	1.0	μS
$t_{OPEN}$ , Break-Before-Make	+25°C	.025	0.1		0.25	0.1		μS
$t_{ON}$ , (EN, $\overline{WR}$ ), Enable Delay (ON)	+25°C			1.0			1.0	μS
$t_{OFF}$ , (EN, $\overline{RS}$ ), Enable and Reset Delay (OFF)	+25°C		0.5	1.0		0.5	1.0	μS
Settling Time (±0.1%)	+25°C		1.0			1.0		μS
	+25°C		1.75			1.75		μS
Q Charge Injection	+25°C		10			10		pC
OFF Isolation (Note 4)	+25°C		-72			-72		dB
OFF Isolation POWER OFF (Note 5)	+25°C		-56			-56		dB
$C_{S(off)}$ , Channel Input Cap.	+25°C		5			5		pF
$C_{D(off)}$ , Channel Output Cap.								
	+25°C		40			40		pF
	+25°C		20			20		pF
$C_A$ , Digital Input Capacitance	+25°C		5			5		pF
$C_{DS(off)}$ , Input to Output capacitance	+25°C		0.1			0.1		pF
<b>POWER REQUIREMENTS</b>								
$P_D$ , Power Dissipation (Note 6)	Full		60	100		60	100	mW
$I_+$ , Current Pin 1 (Note 6)	Full		3.7	6.0		3.7	6.0	mA
$I_-$ , Current Pin 27 (Note 6)	Full		0.3	0.6		0.3	0.6	mA

- Notes: 1.  $V_{OUT} = \pm 10V$ ,  $I_{out} = -100 \mu A$   
 2.  $\Delta R_{ON} = \frac{R_{ON(Max)} - R_{ON(Min)}}{R_{ON(Avg)}} V_{IN} = \pm 10V$   
 3. Analog Overvoltage = ±25V  
 4.  $V_{EN} = 0.8V$ ,  $R_L = 1K\Omega$ ,  $C_L = 50pF$

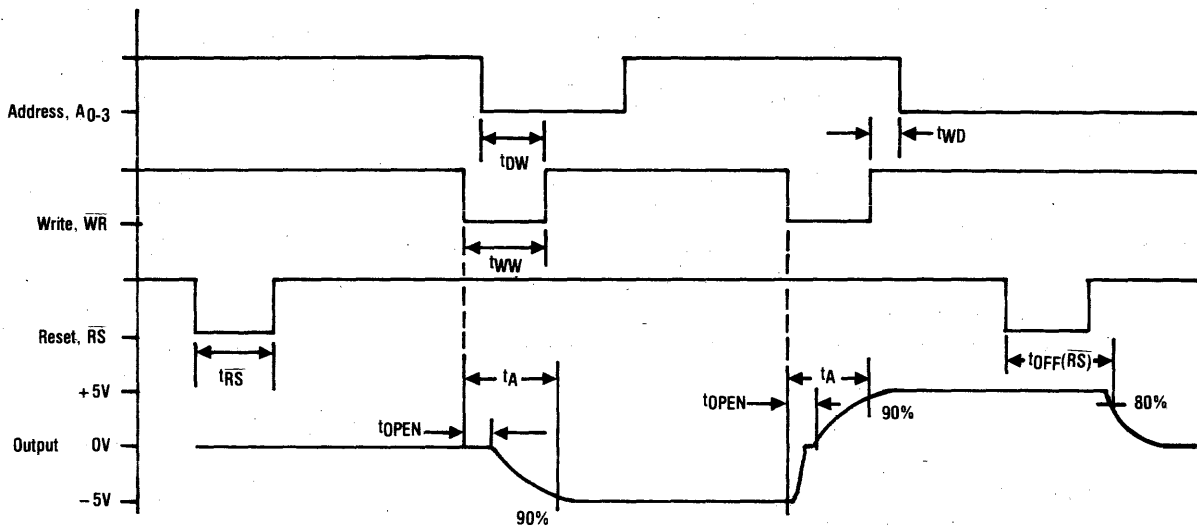
- $V_S = 3V_{rms}$ ,  $f = 500 kHz$   
 Off isolation =  $20 \log \frac{|V_D|}{|V_S|}$   
 5.  $V_+ = 0V$ ,  $V_- = 0V$ ,  $R_L = 1K\Omega$ ,  
 $C_L = 50pF$ ,  $V_S = 3V_{rms}$ ,  $f = 500 kHz$ .  
 6. See Test Circuit #8 for high toggle frequency applications.



### Minimum Timing Requirements

PARAMETER	MEASURED TERMINAL	MIN LIMITS FULL TEMP RANGE	UNITS
$t_{WW}$ , Write Pulse Width	$\overline{WR}$	300	nS
$t_{DW}$ , A, EN Data Valid To WRITE (Stabilization Time)	A0, A1, A2, A3, EN	225	nS
$t_{WD}$ , A, EN Data Valid To WRITE (hold Time)	$\overline{WR}$ $\overline{WR}$	100	nS
$t_{RS}$ , RESET pulse width	A0, A1, A2, A3, EN $\overline{RS}$	400	nS

### Timing Requirements



1. Inputs to HI-506L (507L) : Channel 1 = -5V; Channel 16, (8) = +5V.
2. +VSUPPLY = +15V; -VSUPPLY = -15V.
3. Logic Levels: VAL = 0V; VAH = +2.0V.
4. Time intervals are measured between 50% levels unless otherwise noted.
5. Minimum values for  $t_{RS}$ ,  $t_{DW}$ ,  $t_{WW}$  and  $t_{WD}$  are guaranteed separately but not simultaneously.

Figure 1

### SCHEMATIC DIAGRAM

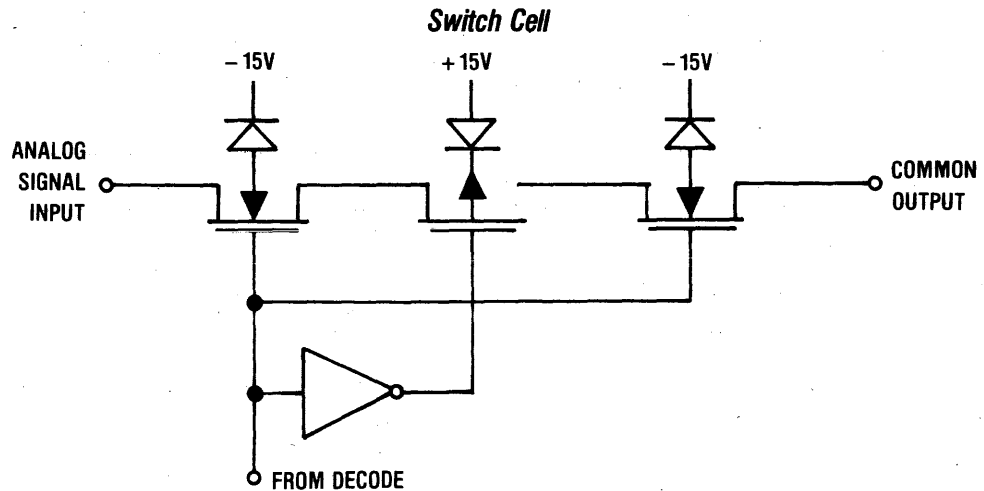


Figure 2



506L

A3	A2	A1	A0	EN	WR	RS	OUTPUT - ON CHANNEL
X	X	X	X	L	L	H	None
X	X	X	X	X	H	H	Previous On Channel.
X	X	X	X	X	X	L	None (latches cleared)
L	L	L	L	H	L	H	Channel 1
L	L	L	H	H	L	H	Channel 2
L	L	H	L	H	L	H	Channel 3
L	L	H	H	H	L	H	Channel 4
L	H	L	L	H	L	H	Channel 5
L	H	L	H	H	L	H	Channel 6
L	H	H	L	H	L	H	Channel 7
L	H	H	H	H	L	H	Channel 8
H	L	L	L	H	L	H	Channel 9
H	L	L	H	H	L	H	Channel 10
H	L	H	L	H	L	H	Channel 11
H	L	H	H	H	L	H	Channel 12
H	H	L	L	H	L	H	Channel 13
H	H	L	H	H	L	H	Channel 14
H	H	H	L	H	L	H	Channel 15
H	H	H	H	H	L	H	Channel 16

507L

A2	A1	A0	EN	WR	RS	OUTPUT - ON CHANNEL PAIR
X	X	X	L	L	H	None
X	X	X	X	H	H	Previous On Channel.
X	X	X	X	X	L	None (latches cleared)
L	L	L	H	L	H	Channel 1A and 1B
L	L	H	H	L	H	Channel 2A and 2B
L	H	L	H	L	H	Channel 3A and 3B
L	H	H	H	L	H	Channel 4A and 4B
H	L	L	H	L	H	Channel 5A and 5B
H	L	H	H	L	H	Channel 6A and 6B
H	H	L	H	L	H	Channel 7A and 7B
H	H	H	H	L	H	Channel 8A and 8B

DESCRIPTION AND APPLICATION

The switch cell of the HI-506L/507L has a different structure than earlier Harris designs (HI-506, HI506A). The new switch (Figure 2) consists of an N-channel, P-channel and N-channel MOSFET in series, as opposed to the transmission gate configuration with an N and P-channel device in parallel. The series N-P-N switch offers higher Off Isolation with power off, and better fault performance. Channel overvoltage protection is inherent since one of the three MOSFETs turn off in the presence of overvoltage. this turn-off process begins well below the supply rail so the  $V_{IN}$  range is less than the power supply range. Electrical performance is guaranteed to  $\pm 10V$  for each channel, and the usable range extends above  $\pm 11$  Volts.

The address inputs  $A_0$ ,  $A_1$ ,  $A_2$ ,  $A_3$ , and ENABLE are latched into an internal buffer when  $WR$  goes high. Each latch output is level shifted into the decode section, which activates the appropriate channel. The device may be reset (all channels OFF) by taking  $RS$  low. Usually,  $RS$  is tied to the system RESET line, to assure that all channels are OFF following a turn-on of power. The reset function overrides all others, just as  $WR$  overrides the address inputs ( $A_0$ - $A_3$  and  $EN$  are ignored when  $WR$  is high). With  $WR$  low and  $RS$  high, the switches respond immediately to a change in channel address; i.e. the latches are "transparent". Refer to Figure 1.



## Single 8/Differential 4 Channel CMOS Analog Multiplexer

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• FAST ACCESS <span style="float: right;">220ns</span></li> <li>• FAST SETTLING (0.01%) <span style="float: right;">600ns</span></li> <li>• LOW R<sub>ON</sub> <span style="float: right;">180 Ω</span></li> <li>• BREAK-BEFORE-MAKE SWITCHING</li> <li>• NO LATCH-UP</li> <li>• TTL/CMOS COMPATIBLE <span style="float: right;">2.4V (LOGIC "1")</span></li> </ul>	<p>These monolithic CMOS multiplexers each include an array of eight analog switches, a digital decode circuit for channel selection, a voltage reference for logic thresholds, and an ENABLE input for device selection when several multiplexers are present.</p> <p>The Dielectric Isolation (DI) process used in fabrication of these devices eliminates the problem of latch-up. Also, DI offers much lower substrate leakage and parasitic capacitance than conventional junction-isolated CMOS (see Application Note 521). Combined with the low ON resistance (180Ω typical), these benefits allow low static error, fast channel switching rates, and fast settling.</p> <p>Switches are guaranteed to break-before-make, so two channels are never shorted together.</p> <p>The switching threshold for each digital input is established by an internal +5V reference, providing a guaranteed min. 2.4V for "1" and max. 0.8V for "0". This allows direct interface without pull-up resistors to signals from most logic families: CMOS, TTL, DTL, and some PMOS. For protection against transient overvoltage, the digital inputs include a series 200Ω resistor and a diode clamp to each supply.</p> <p>The HI-508 is an eight channel single-ended multiplexer, and the HI-509 is a four channel differential version. The recommended supply voltage is ±15V; however, reasonable performance is available down to ±7V. Each device is packaged in a 16 pin DIP.</p> <p>The HI-508/509-5 are specified for operation from 0°C to 70°C. The "-2" versions are specified from -55°C to ±125°C. "Dash 8" (-8) designates -2 parts which have been screened per MIL-STD-883/Method 5004/Class B.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• PRECISION INSTRUMENTS</li> <li>• DATA ACQUISITION SYSTEMS</li> <li>• TELEMETRY</li> </ul>	
PINOUTS	FUNCTIONAL DIAGRAMS
<p><b>HI-508</b></p> <p style="text-align: center;">TOP VIEW</p>	<p><b>HI-508</b></p>
<p><b>HI-509</b></p> <p style="text-align: center;">TOP VIEW</p>	<p><b>HI-509</b></p>



# SPECIFICATIONS

INTERFACE

Harris Semiconductor

## ABSOLUTE MAXIMUM RATINGS (Note 1)

$V_{Supply(+)} \text{ to } V_{Supply(-)}$	40V	Power Dissipation *	750mW
$V_{Supply(+)} \text{ to GND}$	20V	Operating Temperature Ranges:	
$V_{Supply(-)} \text{ to GND}$	20V	HI-508/509-2, -8	-55°C to +125°C
Digital Input Overvoltage:		HI-508/509-5, -6	0°C to 70°C
		HI-508/509-1	-55°C to +200°C
$V_{EN}, V_A \begin{cases} V_{Supply(+)} \\ V_{Supply(-)} \end{cases}$	+4V -4V	Storage Temperature Range	-65°C to +150°C
Analog Input Overvoltage (Note 6):			
$V_D, V_S \begin{cases} V_{Supply(+)} \\ V_{Supply(-)} \end{cases}$	+2V -2V		

\*Derate 9.6mW/°C above  $T_A = 95^\circ\text{C}$

## ELECTRICAL CHARACTERISTICS Unless otherwise specified: Supplies = $\pm 15\text{V}$ , GND = 0V

PARAMETER	TEMP	HI-508/HI-509-2 -55°C to +125°C			HI-508/HI-509-5 0°C to +70°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
$V_S$ , Analog Signal Range	Full	-15		+15	-15		+15	V
RON, On Resistance	+25°C		180	300		180	400	$\Omega$
	Full		230	400		230	500	$\Omega$
$\Delta$ RON, Any Two Channels	+25°C		5			5		%
IS(OFF), Off Input Leakage Current (Note 2)	+25°C			10			10	nA
	Full			50			50	nA
ID(OFF), Off Output Leakage Current	+25°C							
HI-508	Full		10	200		10	200	nA
	Full		10	100		10	100	nA
HI-509	Full		10	200		10	200	nA
	Full		10	100		10	100	nA
ID(ON), On Channel Leakage Current	+25°C							
HI-508	Full		10	200		10	200	nA
	Full		10	100		10	100	nA
HI-509	Full		10	200		10	200	nA
	Full		10	100		10	100	nA
IDIFF, Differential Off Output Leakage Current (HI-509 Only)	+25°C		1	5		1	5	nA
	Full		5	50		5	50	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
$V_{AH}$ , High Threshold	Full	2.4			2.4			V
$V_{AL}$ , Low Threshold	Full			0.8			0.8	V
$I_A$ , Input Leakage Current (High or Low) (Note 3)	Full			1			1	$\mu\text{A}$
<b>SWITCHING CHARACTERISTICS</b>								
$t_A$ , Access (Transition) Time	+25°C		220	500		220	1000	ns
	Full			1000				ns
$t_{OPEN}$ , Break-Before-Make Interval	+25°C		70			70		ns
$t_{ON(EN)}$ , Enable Turn-On	+25°C		210			210		ns
$t_{OFF(EN)}$ , Enable Turn-Off	+25°C		180			180		ns
$t_S$ , Settling Time to 0.1% to 0.01%	+25°C		360			360		ns
	+25°C		600			600		ns
Off Isolation (Note 4)	+25°C		68			68		dB
$C_S(OFF)$ , Channel Input Capacitance	+25°C		5			5		pF
$C_D(OFF)$ , Channel Output Capacitance	+25°C		21			21		pF
$C_A$ , Digital Input Capacitance	+25°C		3			3		pF
$C_{DS(OFF)}$ , Input to Output Capacitance	+25°C		.08			.08		pF
<b>POWER REQUIREMENTS</b>								
$I_+$ , Positive Supply Current (Note 5)	Full			2			2	mA
$I_-$ , Negative Supply Current (Note 5)	Full			1			1	mA
$P_D$ , Power Dissipation	Full			45			45	mW

- NOTES: 1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. Ten nanoamps is the practical limit for high speed measurement in the production test environment. Actually,  $I_S$  (off) is below 100pA for most devices, at 25°C.
3. Digital input leakage is primarily due to the clamp diodes (see Schematic). Typical leakage is less than 1nA at 25°C.

4.  $V_{EN} = 0.8\text{V}$ ,  $R_1 = 1\text{K}$ ,  $C_1 = 15\text{pF}$ ,  $V_S = 7V_{RMS}$ ,  $f = 500\text{kHz}$ . Worst case isolation occurs on channel 4 (HI-508) and channels 4, 8 (HI-509), due to proximity of the output pins.
5.  $V_{EN} = 0\text{V}$  or  $5\text{V}$ . All  $V_A = 0$ .
6. If an overvoltage condition is anticipated (analog input exceeds either power supply voltage), the HARRIS HI-508A/509A multiplexers are recommended.

## TRUTH TABLES

### HI-508

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	"ON" CHANNEL
X	X	X	L	NONE
L	L	L	H	1
L	L	H	H	2
L	H	L	H	3
L	H	H	H	4
H	L	L	H	5
H	L	H	H	6
H	H	L	H	7
H	H	H	H	8

### HI-509

A <sub>1</sub>	A <sub>0</sub>	EN	"ON" CHANNEL
X	X	L	NONE
L	L	H	1
L	H	H	2
H	L	H	3
H	H	H	4



# HI-508A/509A

## 8 Channel CMOS Analog Multiplexers with Overvoltage Protection

INTERFACE

Harris Semiconductor

FEATURES	DESCRIPTION																
<ul style="list-style-type: none"> <li>ANALOG/DIGITAL OVERVOLTAGE PROTECTION</li> <li>FAIL SAFE WITH POWER LOSS (NO LATCHUP)</li> <li>BREAK-BEFORE-MAKE SWITCHING</li> <li>DTL/TTL AND CMOS COMPATIBLE</li> <li>ANALOG SIGNAL RANGE <span style="float: right;">±15V</span></li> <li>ACCESS TIME (TYP.) <span style="float: right;">500ns</span></li> <li>SUPPLY CURRENT AT 1MHz ADDRESS TOGGLE (TYP.) <span style="float: right;">4mA</span></li> <li>STANDBY POWER (TYP.) <span style="float: right;">7.5mW</span></li> </ul>	<p>The HI-508A and HI-509A are dielectrically isolated CMOS analog multiplexers incorporating an important feature; they withstand analog input voltages much greater than the supplies. This is essential in any system where the analog inputs originate outside the equipment. They can withstand a continuous input up to 10 volts greater than either supply, which eliminates the possibility of damage when supplies are off, but input signals are present. Equally important, they can withstand brief input transient spikes of several hundred volts; which otherwise would require complex external protection networks. Necessarily, ON resistance is somewhat higher than similar unprotected devices, but very low leakage currents combine to produce low errors. Application Notes 520 and 521 further explain these features.</p> <p>The HI-508A-2 and HI-509A-2 are specified over -55°C to +125°C while the -5 versions are specified over 0°C to +75°C.</p>																
APPLICATIONS																	
<ul style="list-style-type: none"> <li>DATA ACQUISITION</li> <li>INDUSTRIAL CONTROLS</li> <li>TELEMETRY</li> </ul>																	
PINOUT	FUNCTIONAL DIAGRAM																
<p><b>HI-508A</b></p> <p style="text-align: center;">TOP VIEW</p> <table border="0" style="width: 100%; border-collapse: collapse;"> <tr> <td style="width: 50%;">A<sub>0</sub> 1</td> <td style="width: 50%; text-align: right;">16 A<sub>1</sub></td> </tr> <tr> <td>E<sub>N</sub> 2</td> <td style="text-align: right;">15 A<sub>2</sub></td> </tr> <tr> <td>-V<sub>sup</sub> 3</td> <td style="text-align: right;">14 GND</td> </tr> <tr> <td>IN1 4</td> <td style="text-align: right;">13 +V<sub>sup</sub></td> </tr> <tr> <td>IN2 5</td> <td style="text-align: right;">12 IN5</td> </tr> <tr> <td>IN3 6</td> <td style="text-align: right;">11 IN6</td> </tr> <tr> <td>IN4 7</td> <td style="text-align: right;">10 IN7</td> </tr> <tr> <td>OUT 8</td> <td style="text-align: right;">9 IN8</td> </tr> </table>	A <sub>0</sub> 1	16 A <sub>1</sub>	E <sub>N</sub> 2	15 A <sub>2</sub>	-V <sub>sup</sub> 3	14 GND	IN1 4	13 +V <sub>sup</sub>	IN2 5	12 IN5	IN3 6	11 IN6	IN4 7	10 IN7	OUT 8	9 IN8	<p><b>HI-508A</b></p>
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<p><b>HI-509A</b></p> <p style="text-align: center;">TOP VIEW</p> <table border="0" style="width: 100%; border-collapse: collapse;"> <tr> <td style="width: 50%;">A<sub>0</sub> 1</td> <td style="width: 50%; text-align: right;">16 A<sub>1</sub></td> </tr> <tr> <td>E<sub>N</sub> 2</td> <td style="text-align: right;">15 GND</td> </tr> <tr> <td>-V<sub>sup</sub> 3</td> <td style="text-align: right;">14 +V<sub>sup</sub></td> </tr> <tr> <td>IN1A 4</td> <td style="text-align: right;">13 IN1B</td> </tr> <tr> <td>IN2A 5</td> <td style="text-align: right;">12 IN2B</td> </tr> <tr> <td>IN3A 6</td> <td style="text-align: right;">11 IN3B</td> </tr> <tr> <td>IN4A 7</td> <td style="text-align: right;">10 IN4B</td> </tr> <tr> <td>OUTA 8</td> <td style="text-align: right;">9 OUTB</td> </tr> </table>	A <sub>0</sub> 1	16 A <sub>1</sub>	E <sub>N</sub> 2	15 GND	-V <sub>sup</sub> 3	14 +V <sub>sup</sub>	IN1A 4	13 IN1B	IN2A 5	12 IN2B	IN3A 6	11 IN3B	IN4A 7	10 IN4B	OUTA 8	9 OUTB	<p><b>HI-509A</b></p>
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IN4A 7	10 IN4B																
OUTA 8	9 OUTB																



# SPECIFICATIONS

INTERFACE

## ABSOLUTE MAXIMUM RATINGS

Voltage between Supply Pins	40V	Total Power Dissipation*	725 mW
V+ to Ground	20V	Operating Temperature:	
V <sub>EN</sub> , V <sub>A</sub> , Digital Input Overvoltage:		HI-508A/HI-509A-2	-55°C to +125°C
V <sub>A</sub>   V <sub>Supply</sub> (+) +4V		HI-508A/HI-509A-5	0°C to +75°C
V <sub>A</sub>   V <sub>Supply</sub> (-) -4V		Storage Temperature	-65°C to +150°C
Analog Input Overvoltage:			
V <sub>S</sub>   V <sub>Supply</sub> (+) +20V			
V <sub>S</sub>   V <sub>Supply</sub> (-) -20V			

\*Derate 9.6mW/°C above T<sub>A</sub> = 95°C

## ELECTRICAL CHARACTERISTICS (Unless Otherwise Specified)

Supplies = +15V, -15V; V<sub>AH</sub> (Logic Level High) = +4.0V; V<sub>AL</sub> (Logic Level Low) = +0.8V  
For Test Conditions, consult Performance Characteristics section.

PARAMETER	TEMP.	HI-508A/509A-2 -55°C to +125°C			HI-508A/509A-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
*V <sub>S</sub> , Analog Signal Range	Full	-15		+15	-15		+15	V
*R <sub>ON</sub> , On Resistance (Note 1)	+25°C		1.2	1.5		1.5	1.8	KΩ
	Full		1.5	1.8		1.8	2.0	KΩ
*I <sub>S(OFF)</sub> , Off Input Leakage Current	+25°C		0.03			0.03		nA
	Full			±50			±50	nA
*I <sub>D(OFF)</sub> , Off Output Leakage Current	+25°C		1.0			1.0		nA
	HI-508A Full			±250			±250	nA
HI-509A Full				±125			±125	nA
*I <sub>D(OFF)</sub> with Input Overvoltage Applied (Note 2)	+25°C		4.0			4.0		nA
	Full			2.0				μA
*I <sub>D(ON)</sub> , On Channel Leakage Current	+25°C		0.1			0.1		nA
	HI-508A Full			±250			±250	nA
HI-509A Full				±125			±125	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>AL</sub> , Input Low Threshold	Full			0.8			0.8	V
V <sub>AH</sub> , Input High Threshold (Note 6)	Full	4.0			4.0			V
*I <sub>A</sub> , Input Leakage Current (High or Low)	Full			1.0			1.0	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>A</sub> , Access Time	+25°C		0.5	1.0		0.5		μs
t <sub>OPEN</sub> , Break - Before Make Delay	+25°C		80			80		ns
t <sub>ON(EN)</sub> , Enable Delay (ON)	+25°C		300			300		ns
t <sub>OFF(EN)</sub> , Enable Delay (OFF)	+25°C		300			300		ns
Settling Time (0.1%)	+25°C		1.2			1.2		μs
	(0.025%)	+25°C		3.5		3.5		μs
"OFF Isolation" (Note 3)	+25°C		65			65		dB
C <sub>S</sub> (OFF), Channel Input Capacitance	+25°C		5			5		pF
C <sub>D</sub> (OFF), Channel Output Capacitance	HI-508A +25°C		25			25		pF
	HI-509A +25°C		12			12		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		5			5		pF
C <sub>DS</sub> (OFF), Input to Output Capacitance	+25°C		0.1			0.1		pF
<b>POWER REQUIREMENTS</b>								
P <sub>D</sub> , Power Dissipation	Full		7.5			7.5		mW
*I <sub>+</sub> , Current (Note 4)	Full		0.5	2.0		0.5	5.0	mA
*I <sub>-</sub> , Current (Note 4)	Full		0.02	1.0		0.02	2.0	mA
*I <sub>+</sub> , Standby (Note 5)	Full		0.5	2.0		0.5	5.0	mA
*I <sub>-</sub> , Standby (Note 5)	Full		0.02	1.0		0.02	2.0	mA

- NOTES: 1. V<sub>OUT</sub> = ±10V, I<sub>OUT</sub> = -100 μA  
 2. Analog Overvoltage = ±33V  
 3. V<sub>EN</sub> = 0.8V, R<sub>L</sub> = 1K, C<sub>L</sub> = 7pF, V<sub>S</sub> = 3V RMS, f = 500KHz  
 4. V<sub>EN</sub> = +4.0V  
 5. V<sub>EN</sub> = 0.8V  
 6. To drive from DTL/TTL Circuits, 1KΩ pull-up resistors to +5.0V supply are recommended

\* 100% Tested for Dash 8 at +25°C and +125°C Only.

## TRUTH TABLES

### HI-508A

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	"ON" CHANNEL
X	X	X	L	NONE
L	L	L	H	1
L	L	H	H	2
L	H	L	H	3
L	H	H	H	4
H	L	L	H	5
H	L	H	H	6
H	H	L	H	7
H	H	H	H	8

### HI-509A

A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH PAIR
X	X	L	NONE
L	L	H	1
L	H	H	2
H	L	H	3
H	H	H	4

Harris Semiconductor

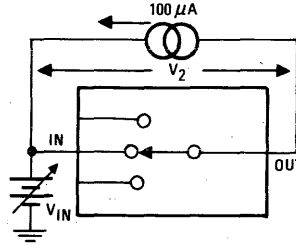
# PERFORMANCE CHARACTERISTICS AND TEST CIRCUITS



UNLESS OTHERWISE SPECIFIED:  $T_A = 25^\circ\text{C}$ ,  $V_{\text{SUPPLY}} = +15\text{V}$ ,  $V_{\text{AH}} = +4\text{V}$ ,  $V_{\text{AL}} = 0.8\text{V}$

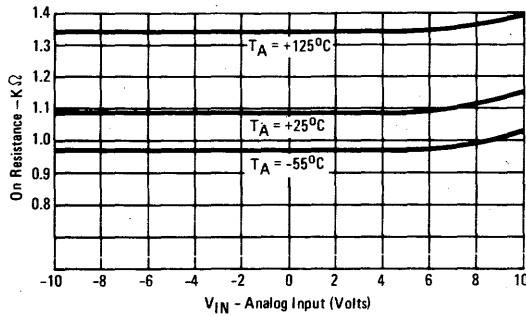
**TEST CIRCUIT NO. 1**

$$R_{\text{ON}} = \frac{V_2}{100 \mu\text{A}}$$

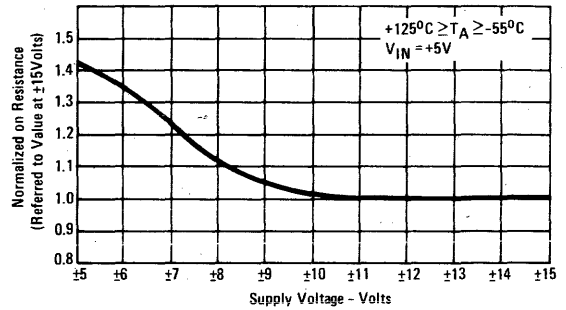


**ON RESISTANCE vs. INPUT SIGNAL LEVEL, SUPPLY VOLTAGE**

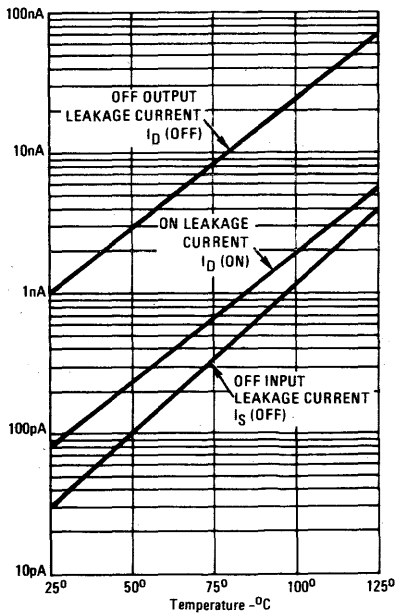
**ON RESISTANCE vs. ANALOG INPUT VOLTAGE**



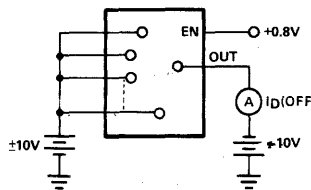
**NORMALIZED ON RESISTANCE vs. SUPPLY VOLTAGE**



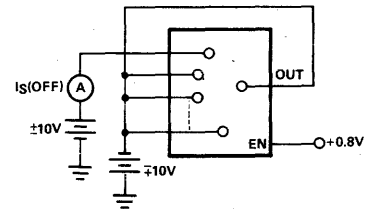
**LEAKAGE CURRENT vs. TEMPERATURE**



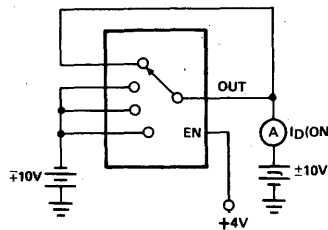
**TEST CIRCUIT NO. 2\***



**TEST CIRCUIT NO. 3\***

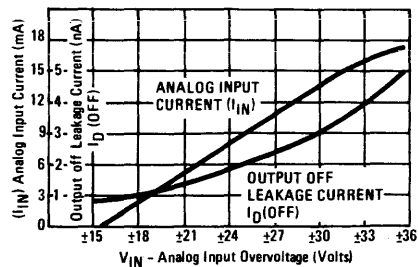


**TEST CIRCUIT NO. 4\***



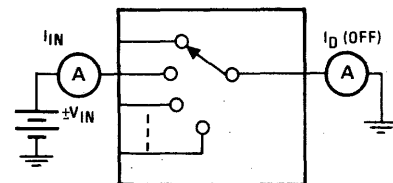
\*Two measurements per channel: +10V/-10V and -10V/+10V. (Two measurements per device for ID(OFF): +10V/-10V and -10V/+10V.)

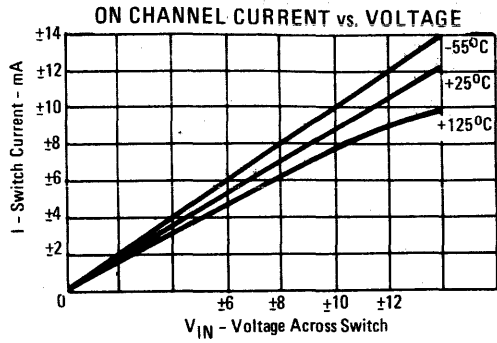
**ANALOG INPUT OVERVOLTAGE CHARACTERISTICS**



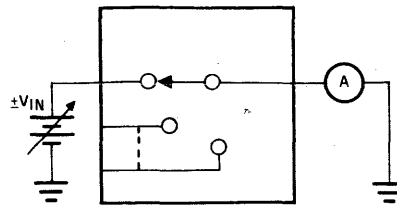
**ANALOG INPUT OVERVOLTAGE CHARACTERISTICS**

**TEST CIRCUIT NO. 5**

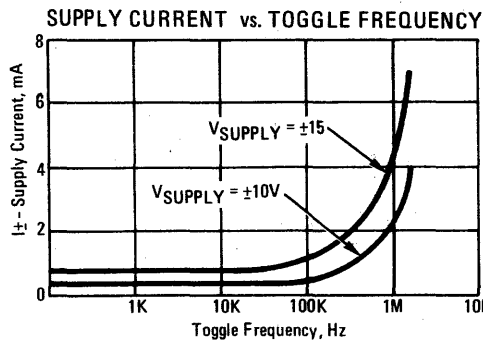




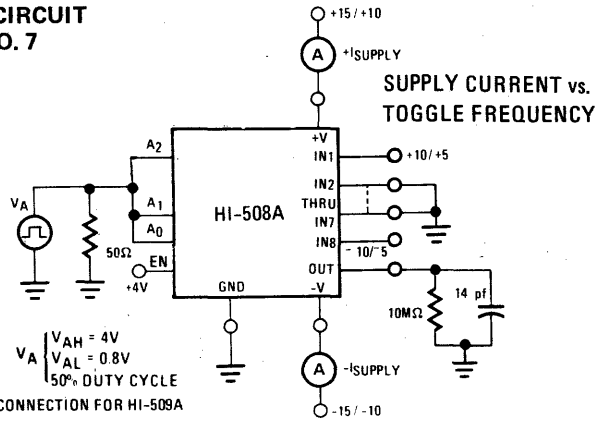
**TEST CIRCUIT NO. 6**



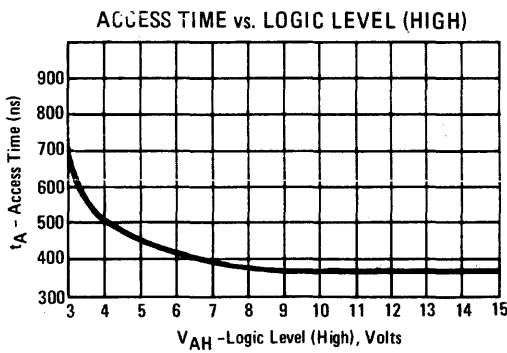
**ON CHANNEL CURRENT vs. VOLTAGE**



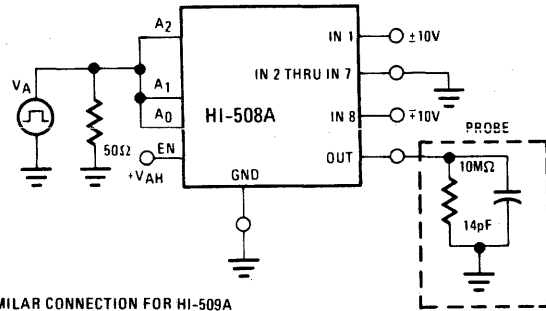
**TEST CIRCUIT NO. 7**



\*SIMILAR CONNECTION FOR HI-509A

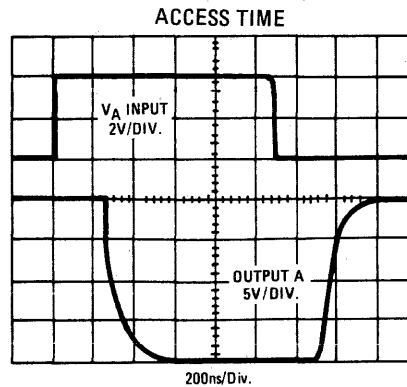
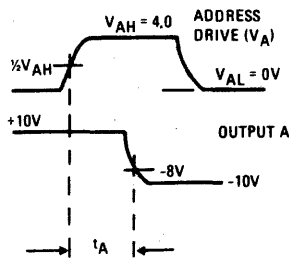


**TEST CIRCUIT NO. 8**



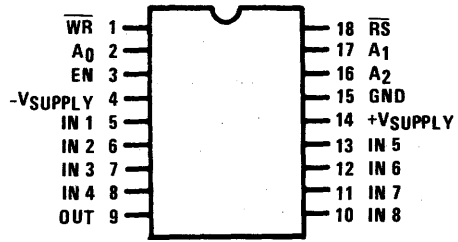
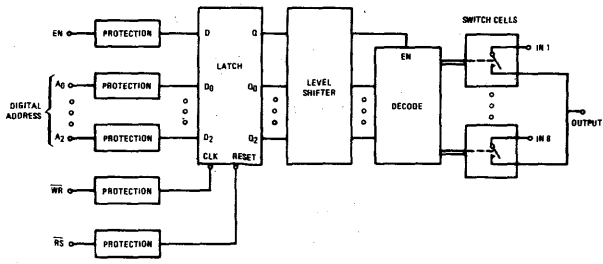
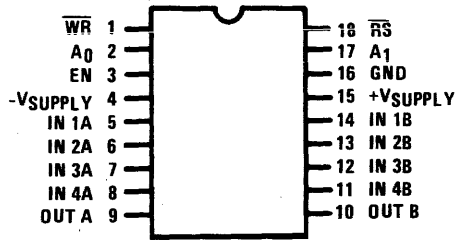
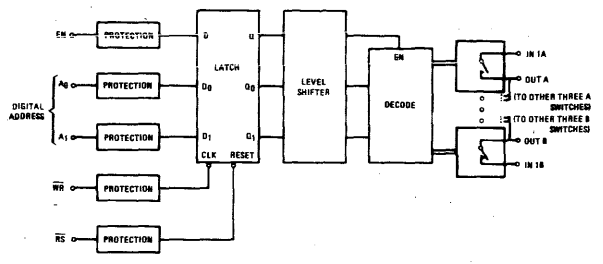
\*SIMILAR CONNECTION FOR HI-509A

**SWITCHING WAVEFORMS**



# HI-508L/HI-509L

## Single 8/Differential 4 Channel CMOS Analog Multiplexers With Latches And Overvoltage Protection

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• Analog Overvoltage protection</li> <li>• Resettable Latches (<math>\overline{RS}</math>)</li> <li>• TTL/DTL and CMOS Compatible</li> <li>• Failsafe for conditions of Overvoltage &amp; Loss of Power</li> <li>• No SCR Latch-up</li> <li>• Break-before-make switching</li> <li>• Microprocessor Bus compatible</li> <li>• Very low leakage - <math>I_{D(off)} \leq 8nA</math> (typ) over full temp range</li> <li>• Access time - <math>t_A = 500nS</math> (typ)</li> <li>• Minimum write pulse width (<math>\overline{WR}</math>) = 300 nS</li> <li>• OFF isolation = -100dB, typ @ 10kHz</li> </ul>	<p>These monolithic CMOS multiplexers feature on-board address latches, plus overvoltage protection for the analog inputs and the output as well. Each mode accommodates digital inputs for channel selection and an Enable input for device selection under program control. In addition, Write (<math>\overline{WR}</math>) and Reset (<math>\overline{RS}</math>) inputs allow the program to store or clear the channel address.</p> <p>The overvoltage performance of these multiplexers is particularly useful in redundant systems, where the inputs and output should present a very high impedance when power is off. This is achieved by a switch cell with three MOSFET's in series, rather than the conventional single transmission gate design.</p> <p>Each channel can withstand a DC overvoltage up to a magnitude of 10VDC greater than the power supply (or to <math>\pm 25VDC</math> with power off). An OFF channel remains OFF for this condition. If the channel is ON, the output voltage is clamped below the supply rail, which protects the load circuit.</p> <p>The HI-508L offers 8 single-ended channels, and the HI-509L is a 4 channel differential version. The recommended supply voltages are <math>\pm 15V</math>, through operation at reduced levels or with a single supply may also be implemented. The package is an 18 pin ceramic or epoxy DIP.</p> <p>Each product is specified for the commercial temperature range (0°C to +75°C, -5 suffix) and the military range (-55°C to +125°C, -2 suffix). Military parts screened per MIL-STD-883 method 5004, Class B are offered with a "-8" suffix.</p>
PINOUT	FUNCTIONAL DIAGRAM
<p><b>HI-508L</b></p> 	<p><b>HI-508L</b></p> 
<p><b>HI-509L</b></p> 	<p><b>HI-509L</b></p> 

CAUTION: Electronic devices are sensitive to electrostatic discharge. Proper I.C. handling procedures should be followed.

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**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage Between Pins 4 and 14	44V	Operating Temperature	
Digital Input Overvoltage, $V_A, V_{EN}, \overline{V_{RS}}, \overline{V_{WR}}$			
V supply (+)	+4V	HI-508L/509L-2	-55°C to 125°C
V supply (-)	-4V	HI-508L/509L-5	0°C to 75°C
Analog Overvoltage		Storage Temperature	-65°C to +150°C
Input to Ground	±25VDC		
Total Power Dissipation* (Package)	1200mW	*Derate-8mW/°C above $T_A = +75^\circ\text{C}$	

**ELECTRICAL CHARACTERISTICS** (Unless otherwise specified)

+V supply = 15V, -V supply = -15V,  $V_{AH}$  (Logic High) = 2.0V,  $V_{AL}$  (Logic Low) = 0.8V

PARAMETER	HI-508L/509L-2 -55°C to +125°C				HI-508L/509L-5 0°C to +75°C			UNITS
	TEMP	MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
$V_S$ Analog Signal Range	Full		±10			±10		V
$R_{ON}$ , ON Resistance (Note 1)	+25°C			1.2			1.5	KΩ
	Full			1.8			1.8	KΩ
$\Delta R_{ON}$ , Change In $R_{ON}$ (Note 2) between channels	+25°C		5			5		%
$I_{S(off)}$ , OFF input leakage current	+25°C			10			10	nA
	Full		7	50		7	50	nA
$I_{D(off)}$ , OFF output leakage current	+25°C			10			10	nA
HI-508L	Full		5	100		5	100	nA
HI-509L	Full			50			50	nA
$I_{D(on)}$ , ON Channel leakage current	+25°C		5	10			10	nA
HI-508L	Full		5	100		5	100	nA
HI-509L	Full			50			50	nA
<b>FAULT CHARACTERISTICS</b>								
$I_{S(off)}$ , with Power OFF	Full		10	1000		10	5000	nA
$I_{S(off)}$ , overvoltage (Note 3)	Full		10	750		10	2500	nA
$I_{D(off)}$ , with input overvoltage applied (Note 3)	+25°C		5			5		nA
	Full		10	750		10	2500	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
$V_{AL}$ , Input Low Threshold	Full		1.4	0.8		1.4	0.8	V
$V_{AH}$ , Input High Threshold	Full	2.0	1.4		2.0	1.4		V
$I_{AH}$ , Input High Current	Full		10	40		10	40	μA
$I_{AL}$ , Input Low Current	Full		40	200		40	200	μA
<b>DYNAMIC SWITCHING CHARACTERISTICS</b>								
$t_a$ , Access Time	+25°C		0.5	1.0		0.5	1.0	μS
$t_{OPEN}$ , Break-Before-Make	+25°C	.025	0.1		0.25	0.1		μS
$t_{ON}$ , (EN, $\overline{WR}$ ), Enable Delay (ON)	+25°C			1.0			1.0	μS
$t_{OFF}$ , (EN, $\overline{RS}$ ), Enable and Reset Delay (OFF)	+25°C		0.5	1.0		0.5	1.0	μS
Settling Time (±0.1%)	+25°C		1.0			1.0		μS
(±0.01%)	+25°C		1.75			1.75		μS
Q Charge Injection	+25°C		10			10		pC
OFF Isolation (Note 4)	+25°C		-72			-72		dB
OFF Isolation POWER OFF (Note 5)	+25°C		-56			-56		dB
$C_{S(off)}$ , Channel Input Cap.	+25°C		5			5		pF
$C_{D(off)}$ , Channel Output Cap.								
HI-508L	+25°C		20			20		
HI-509L	+25°C		10			10		pF
$C_A$ , Digital Input Capacitance	+25°C		5			5		pF
$C_{DS(off)}$ , Input to Output capacitance	+25°C		0.1			0.1		pF
<b>POWER REQUIREMENTS</b>								
$P_D$ , Power Dissipation (Note 6)	Full		60	100		60	100	mW
$I_+$ , Current Pin 14 (Note 6)	Full		3.7	6.0		3.7	6.0	mA
$I_-$ , Current Pin 4 (Note 6)	Full		0.3	0.6		0.3	0.6	mA

Notes: 1.  $V_{OUT} = \pm 10V$ ,  $I_{out} = -100 \mu A$

2.  $\Delta R_{ON} = \frac{R_{ON(Max)} - R_{ON(Min)}}{R_{ON(Avg)}} V_{IN} = \pm 10V$

3. Analog Overvoltage = ±25V

4.  $V_{EN} = 0.8V$ ,  $R_L = 1K\Omega$ ,  $C_L = 50pF$

$V_S = 3V_{rms}$ ,  $f = 500 \text{ kHz}$

Off isolation =  $20 \log \left| \frac{V_D}{V_S} \right|$

5.  $V_+, V_- = 0V$ ,  $R_L = 1K\Omega$   
 $C_L = 50pF$ ,  $V_S = 3V_{rms}$ ,  $f = 500 \text{ kHz}$ .

6. See Test Circuit #8 for high toggle frequency applications.

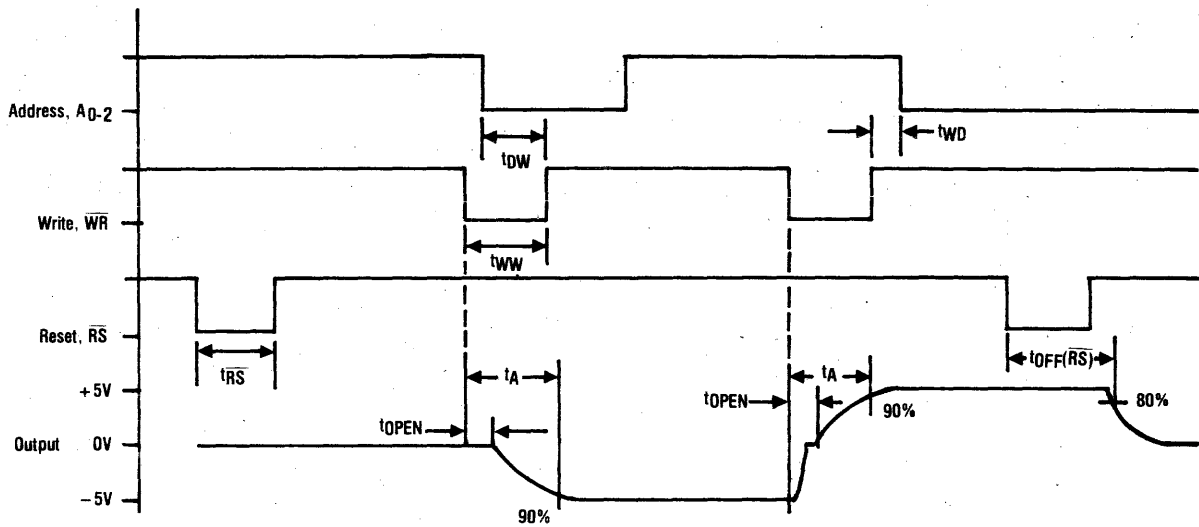




### Minimum Timing Requirements

PARAMETER	MEASURED TERMINAL	MIN LIMITS FULL TEMP RANGE	UNITS
$t_{WW}$ , Write Pulse Width	WR	300	nS
$t_{DW}$ , A, EN Data Valid To WRITE (Stabilization Time)	A0, A1, A2, A3, EN	225	nS
$t_{WD}$ , A, EN Data Valid To WRITE (hold Time)	$\overline{WR}$ A0, A1, A2, A3, EN	100	nS
$t_{RS}$ , RESET pulse width	RS	400	nS

### Timing Requirements



1. Inputs to HI-508L (509L) : Channel 1 = -5V; Channel 8, (4) = +5V.
2. +V<sub>SUPPLY</sub> = +15V; -V<sub>SUPPLY</sub> = -15V.
3. Logic Levels: V<sub>AL</sub> = 0V; V<sub>AH</sub> = +2.0V.
4. Time intervals are measured between 50% levels unless otherwise noted.
5. Minimum values for  $t_{RS}$ ,  $t_{DW}$ ,  $t_{WW}$  and  $t_{WD}$  are guaranteed separately but not simultaneously.

Figure 1

### SCHEMATIC DIAGRAM

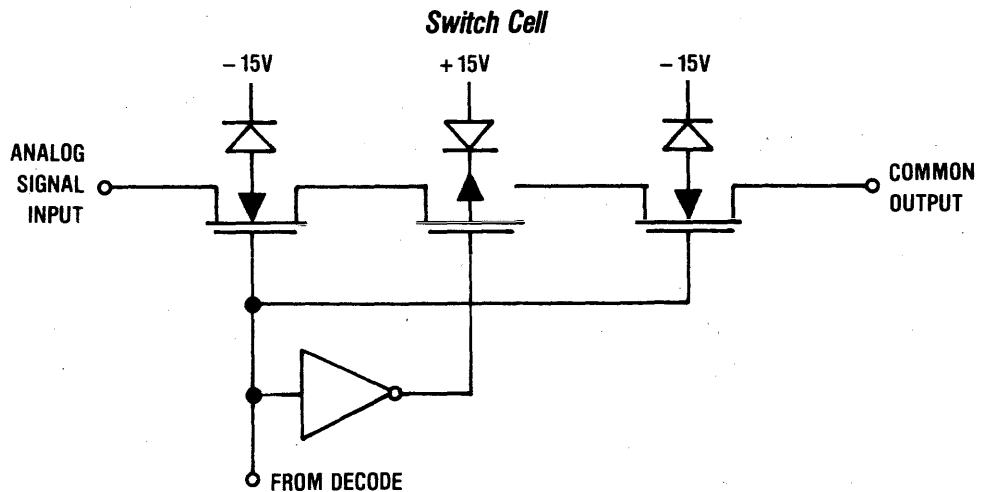


Figure 2

## 508L

A2	A1	A0	EN	$\overline{WR}$	$\overline{RS}$	OUTPUT - ON CHANNEL
X	X	X	L	L	H	None
X	X	X	X	L	H	Previous On Channel.
X	X	X	X	X	L	None (latches cleared)
L	L	L	H	L	H	Channel 1
L	L	H	H	L	H	Channel 2
L	H	L	H	L	H	Channel 3
L	H	H	H	L	H	Channel 4
H	L	L	H	L	H	Channel 5
H	L	H	H	L	H	Channel 6
H	H	L	H	L	H	Channel 7
H	H	H	H	L	H	Channel 8

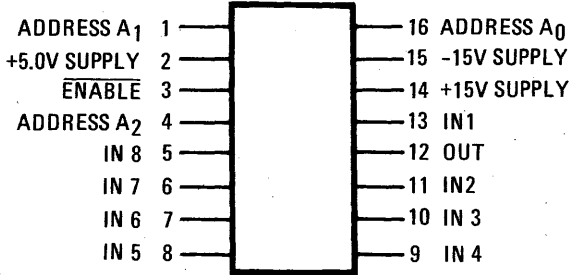
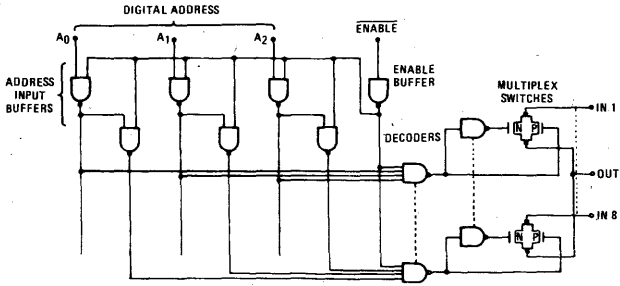
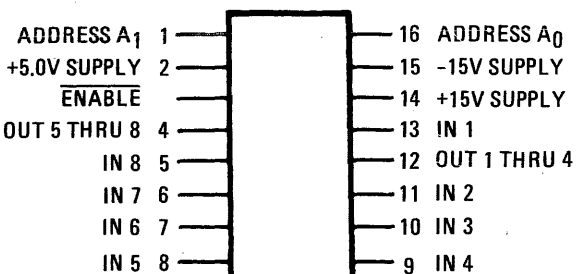
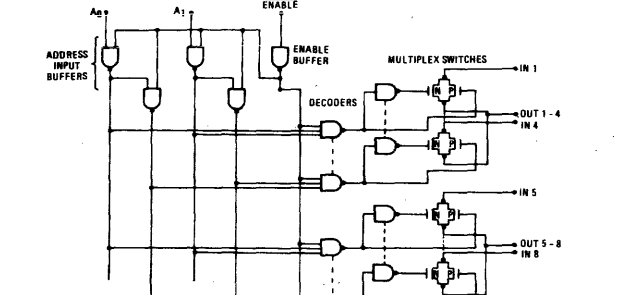
## 507L

A1	A0	EN	$\overline{WR}$	$\overline{RS}$	OUTPUT - ON CHANNEL PAIR
X	X	L	L	H	None
X	X	X	L	H	Previous On Channel.
X	X	X	X	L	None (latches cleared)
L	L	H	L	H	Channel 1A and 1B
L	H	H	L	H	Channel 2A and 2B
H	L	H	L	H	Channel 3A and 3B
H	H	H	L	H	Channel 4A and 4B

**DESCRIPTION AND APPLICATION**

The switch cell of the HI-508L/509L has a different structure than earlier Harris designs (HI-508, HI508A). The new switch (Figure 3) consists of an N-channel, P-channel and N-channel MOSFET in series, as opposed to the transmission gate configuration with an N and P-channel device in parallel. The series N-P-N switch offers higher Off isolation with power off, and better fault performance. Channel overvoltage protection is inherent since one of the three MOSFETs turn off in the presence of overvoltage. This turn-off process begins well below the supply rail so the  $V_{IN}$  range is less than the power supply range. Electrical performance is guaranteed to  $\pm 10V$  for each channel, and the usable range extends above  $\pm 11$  Volts.

The address inputs  $A_0$ ,  $A_1$ ,  $A_2$ , and ENABLE are latched into an internal buffer when  $\overline{WR}$  goes high. Each latch output is level shifted into the decode section, which activates the appropriate channel. The device may be reset (all channels OFF) by taking  $\overline{RS}$  low. Usually,  $\overline{RS}$  is tied to the system RESET line, to assure that all channels are OFF following a turn-on of power. The reset function overrides all others, just as  $\overline{WR}$  overrides the address inputs ( $A_0$ - $A_2$  and EN are ignored when  $\overline{WR}$  is high). With  $\overline{WR}$  low and  $\overline{RS}$  high, the switches respond immediately to a change in channel address; i.e., the latches are "transparent". Refer to Figure 1.

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• SIGNAL RANGE <span style="float: right;">±15V</span></li> <li>• "ON" RESISTANCE (TYP.) <span style="float: right;">250Ω</span></li> <li>• INPUT LEAKAGE AT +125°C (TYP.) <span style="float: right;">20nA</span></li> <li>• ACCESS TIME (TYP.) <span style="float: right;">350ns</span></li> <li>• POWER CONSUMPTION (TYP.) <span style="float: right;">5mW</span></li> <li>• DTL/TTL COMPATIBLE ADDRESS</li> <li>• -55°C to +125°C OPERATION</li> </ul>	<p>The HI-1818A/1828A are monolithic high performance CMOS analog multiplexers offering built-in channel selection decoding plus an inhibit (enable) input for disabling all channels. Dielectric Isolation (DI) processing is used for enhanced reliability and performance (see Application Note 521). Substrate leakage and parasitic capacitance are much lower, resulting in extremely low static errors and high throughput rates. Low output leakage (typically 0.1nA) and low channel ON resistance (250 Ω) assure optimum performance in low level or current mode applications.</p> <p>The 1818A is a single-ended 8 channel multiplexer, while the HI-1828A is a differential 4 channel version. Either device is ideally suited for medical instrumentation, telemetry systems, and microprocessor based data acquisition systems.</p> <p>The HI-1818A-2 and HI-1828A are specified over -55°C to +125°C, while the -5 versions are specified over 0°C to +75°C.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• DATA ACQUISITION SYSTEMS</li> <li>• PRECISION INSTRUMENTATION</li> <li>• DEMULTIPLEXING</li> <li>• SELECTOR SWITCH</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p><b>HI-1818A</b></p> <p style="text-align: center;">Top View</p> 	<p><b>HI-1818A</b></p> 
<p><b>HI-1828A</b></p> <p style="text-align: center;">Top View</p> 	<p><b>HI-1828A</b></p> 



**ABSOLUTE MAXIMUM RATINGS (NOTE 1)**

Supply Voltage Between Pins 14 and 15 **40.0V**  
 Logic Supply Voltage, Pin 2 **30.0V**  
 Analog Input Voltage:  $V_{Supply}^{+}$  **+2V**  
 $V_{Supply}^{-}$  **-2V**

Digital Input Voltage **V-Supply to V+ Supply**  
 Total Power Dissipation (Note 2) **750mW**  
 Storage Temperature Range **-65°C to +150°C**

**ELECTRICAL CHARACTERISTICS**

Supplies = +15V, -15V, +5V

PARAMETER	TEMP.	HI-1818A-2/1828A-2 -55°C to +125°C			HI-1818A-5/1828A-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
* $V_{IN}$ , Analog Signal Range	Full	-15		+15	-15		+15	V
* $R_{ON}$ , ON Resistance (Note 3)	+25°C		250	400		250	400	$\Omega$
	Full		300	500		300	500	$\Omega$
* $I_S$ (OFF), Input Leakage Current	Full		20	50		20	50	nA
* $I_D$ (ON), On Channel Leakage Current (HI-1818A)	Full		100	250		100	250	nA
(HI-1828A)	Full		50	125		50	125	nA
* $I_D$ (OFF) Output Leakage Current (HI-1818A)	Full		100	250		100	250	nA
(HI-1828A)	Full		50	125		50	125	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
$V_{AL}$ , Input Low Threshold	Full			0.4			0.4	V
$V_{AH}$ , Input High Threshold (Note 4)	Full	4.0			4.0			V
$I_{IA}$ , Input Leakage Current	Full		.01	1		.01	1	$\mu$ A
<b>SWITCHING CHARACTERISTICS</b>								
$T_S$ , Access Time (Note 5)	+25°C		350		350			ns
Break-Before-Make Delay	+25°C		100		100			ns
Settling Time (0.1%)	+25°C		1.08		1.08			$\mu$ s
(0.025%)	+25°C		2.8		2.8			$\mu$ s
$C_{IN}$ , Channel Input Capacitance	+25°C		4		4			pF
$C_{OUT}$ , Channel Output Capacitance (HI-1818A)	+25°C		20		20			pF
(HI-1828A)	+25°C		10		10			pF
$C_{DS}$ (OFF), Drain-To-Source Capacitance	+25°C		0.6		0.6			pF
$C_D$ , Digital Input Capacitance	+25°C		5		5			pF
<b>POWER REQUIREMENTS</b>								
$P_D$ , Power Dissipation	Full		5		5			mW
$P_{DS}$ , Standby Power (Note 6)	Full		5		5			mW
* $I_{+}$ , Current Pin 14	Full		0.1	0.5		0.1	1	mA
* $I_{-}$ , Current Pin 15	Full		0.3	1		0.3	2	mA
* $I_L$ , Current Pin 2	Full		0.3	1		0.3	2	mA

**TRUTH TABLES**

HI-1818A				
ADDRESS				"ON" CHANNEL
A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	$\overline{EN}$	
L	L	L	L	1
L	L	H	L	2
L	H	L	L	3
L	H	H	L	4
H	L	L	L	5
H	L	H	L	6
H	H	L	L	7
H	H	H	L	8
X	X	X	H	NONE

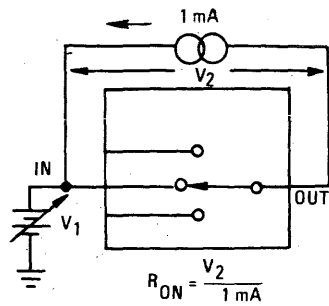
HI-1828A			
ADDRESS			"ON" CHANNELS
A <sub>1</sub>	A <sub>0</sub>	$\overline{EN}$	
L	L	L	1 and 5
L	H	L	2 and 6
H	L	L	3 and 7
H	H	L	4 and 8
X	X	H	NONE

- NOTES: 1. Voltage ratings apply when voltages at all other pins are within their normal operating ranges.  
 2. Derate 12.3 mW/°C above 110°C.  
 3.  $V_{OUT} = \pm 10V$   $I_{OUT} = -1mA$ .  
 4. To drive from DTL/TTL circuits, 1K pull-up resistors to + 5.0V supply are recommended.  
 5. Time measured to 90% of final output level;  $V_{OUT} = -5.0V$  to +5.0V, Digital Inputs = 0V to + 4.0V.  
 6. Voltage at Pin 3, ENABLE = + 4.0V.

\* 100% Tested for Dash 8 at +25°C and +125°C Only.

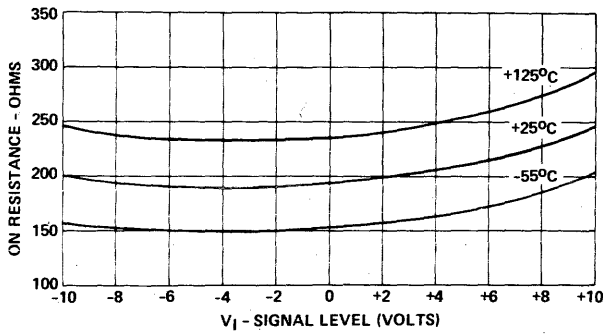


ON RESISTANCE vs ANALOG SIGNAL LEVEL

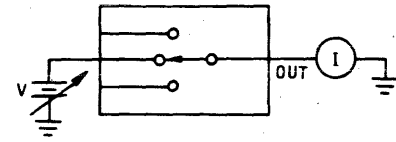


$$R_{ON} = \frac{V_2}{1 \text{ mA}}$$

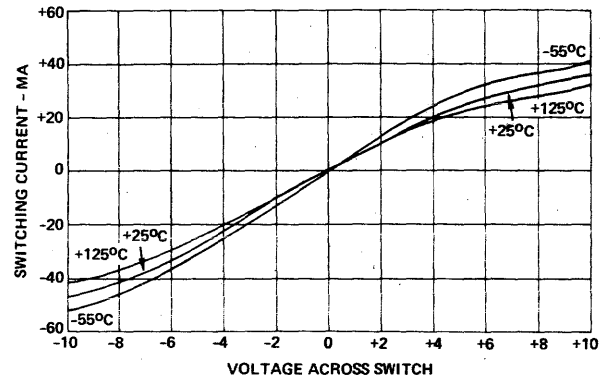
Test Circuit



ON CHANNEL CURRENT vs VOLTAGE

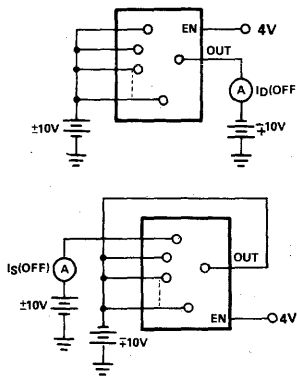


Test Circuit

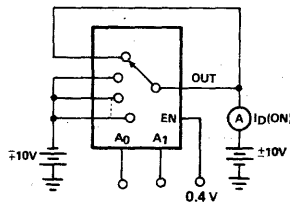


LEAKAGE CURRENTS vs TEMPERATURE

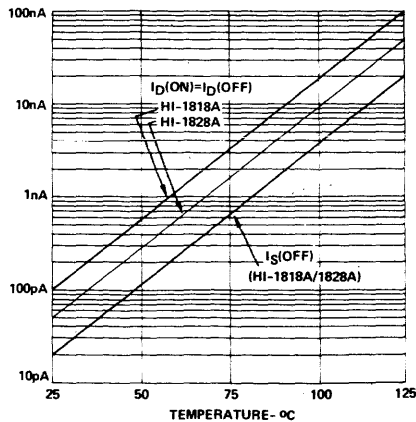
OFF LEAKAGE



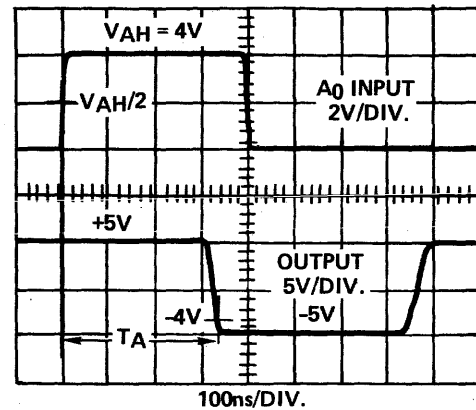
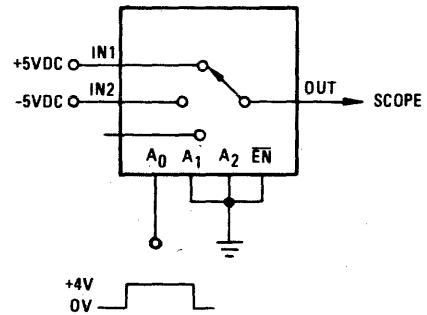
ON LEAKAGE



\*Two measurements per channel:  
+10V/-10V and -10V/+10V.  
†Two measurements per device for I<sub>D</sub>(OFF):  
+10V/-10V and -10V/+10V.

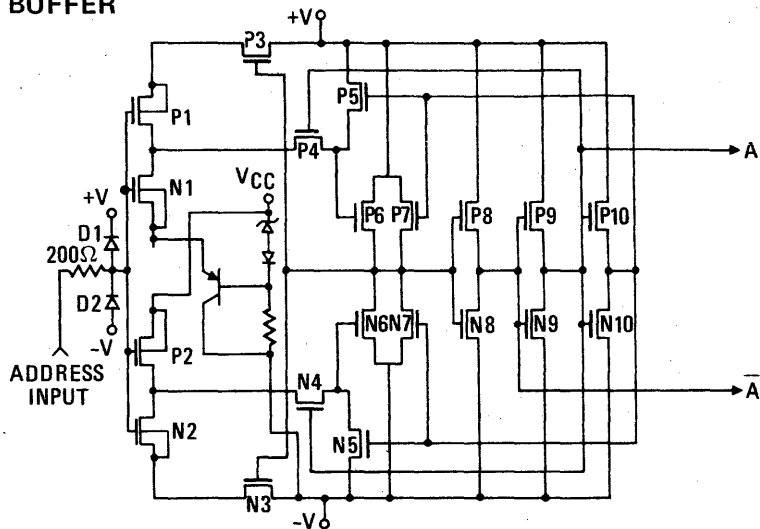


ACCESS TIME



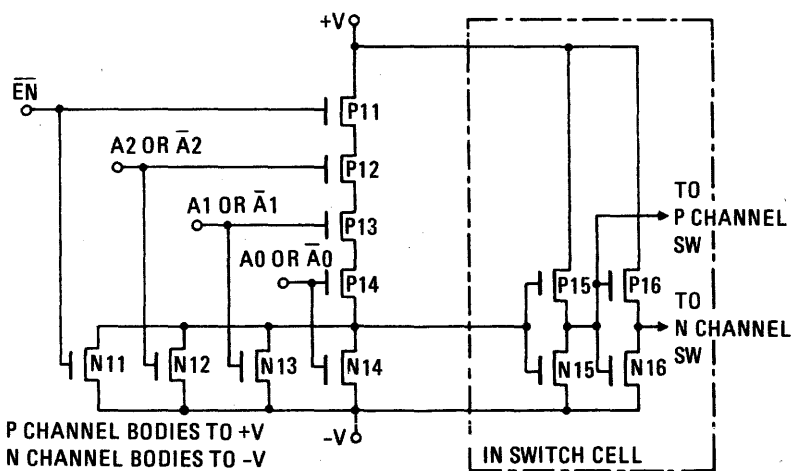


ADDRESS INPUT BUFFER



ALL N-CHANNEL BODIES TO V-  
ALL P-CHANNEL BODIES TO V+ UNLESS OTHERWISE INDICATED.

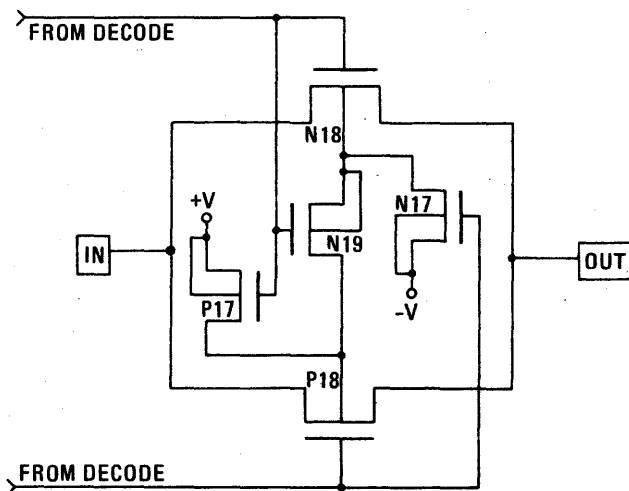
DECODER GATE



P CHANNEL BODIES TO +V  
N CHANNEL BODIES TO -V

A2 OR A2-bar NOT USED FOR HI-1828A

MULTIPLEX SWITCH



# HI-516

## 16 Channel/Differential 8 Channel CMOS High Speed Analog Multiplexer

<b>FEATURES</b>	<b>DESCRIPTION</b>																																																																																				
<ul style="list-style-type: none"> <li>● ACCESS TIME (TYP) <span style="float: right;">100ns</span></li> <li>● SETTLING TIME (TYP TO 0.01%) <span style="float: right;">800ns</span></li> <li>● LOW LEAKAGE <math>I_S</math> OFF <span style="float: right;">10pA</span>  <math>I_D</math> OFF <span style="float: right;">35pA</span></li> <li>● LOW CAPACITANCE <math>C_S</math> OFF <span style="float: right;">2.5pF</span>  <math>C_D</math> OFF <span style="float: right;">18pF</span></li> <li>● HIGH OFF ISOLATION AT 1MHz <span style="float: right;">80dB</span></li> <li>● LOW CHARGE INJECTION <span style="float: right;">0.3pC</span></li> <li>● SINGLE ENDED TO DIFFERENTIAL SELECTABLE (SDS)</li> <li>● LOGIC LEVEL SELECTABLE (LLS)</li> </ul>	<p>The HI-516 is a monolithic dielectrically isolated, high speed, high performance CMOS analog multiplexer. It offers unique built-in channel selection decoding plus an inhibit input for disabling all channels. The dual function of address input <math>A_3</math> enables the HI-516 to be user programmed either as a single ended 16-channel multiplexer, by connecting 'out A' to 'out B' and using <math>A_3</math> as a digital address input, or as an 8-channel differential multiplexer by connecting <math>A_3</math> to the <math>V^-</math> supply. The substrate leakages and parasitic capacitances are reduced substantially using the Harris dielectric isolation process to achieve optimum performances in both high and low level signal applications. The low output leakage current (<math>I_D</math> Off <math>&lt; 100pA @ 25^{\circ}C</math>) and fast settling (<math>t_{SETTLE} = 800ns</math> to 0.01%) characteristics of the device make it an ideal choice for high speed data acquisition systems, precision instrumentation, and industrial process controls.</p> <p>The HI-516 is available in a 28 lead dual-in-line package. HI-516-5 is specified for operation over <math>0^{\circ}C</math> to <math>+75^{\circ}C</math>, and the HI-516-2 over <math>-55^{\circ}C</math> to <math>+125^{\circ}C</math>. Processing to MIL-STD-883A, Class B screening is available by selecting the HI-516-8.</p>																																																																																				
<b>APPLICATIONS</b>	<ul style="list-style-type: none"> <li>● DATA ACQUISITION SYSTEMS</li> <li>● PRECISION INSTRUMENTATION</li> <li>● INDUSTRIAL CONTROL</li> </ul>																																																																																				
<b>PINOUT</b>	<b>FUNCTIONAL DIAGRAM</b>																																																																																				
<p style="text-align: center;">Section 11 for Packaging</p> <p style="text-align: center;">TOP VIEW</p> <table style="width: 100%; border-collapse: collapse;"> <tr> <td style="width: 15%;"><math>V^+</math></td> <td style="width: 10%; border: 1px solid black; text-align: center;">1</td> <td style="width: 10%;"></td> <td style="width: 10%; border: 1px solid black; text-align: center;">28</td> <td style="width: 15%;"></td> <td style="width: 10%; border: 1px solid black; text-align: center;">OUT A</td> </tr> <tr> <td>OUT B</td> <td style="border: 1px solid black; text-align: center;">2</td> <td></td> <td style="border: 1px solid black; text-align: center;">27</td> <td></td> <td style="border: 1px solid black; text-align: center;"><math>V^-</math></td> </tr> <tr> <td>NC</td> <td style="border: 1px solid black; text-align: center;">3</td> <td></td> <td style="border: 1px solid black; text-align: center;">26</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN8/8A</td> </tr> <tr> <td>IN16/8B</td> <td style="border: 1px solid black; text-align: center;">4</td> <td></td> <td style="border: 1px solid black; text-align: center;">25</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN7/7A</td> </tr> <tr> <td>IN15/7B</td> <td style="border: 1px solid black; text-align: center;">5</td> <td></td> <td style="border: 1px solid black; text-align: center;">24</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN6/6A</td> </tr> <tr> <td>IN14/6B</td> <td style="border: 1px solid black; text-align: center;">6</td> <td></td> <td style="border: 1px solid black; text-align: center;">23</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN5/5A</td> </tr> <tr> <td>IN13/5B</td> <td style="border: 1px solid black; text-align: center;">7</td> <td></td> <td style="border: 1px solid black; text-align: center;">22</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN4/4A</td> </tr> <tr> <td>IN12/4B</td> <td style="border: 1px solid black; text-align: center;">8</td> <td></td> <td style="border: 1px solid black; text-align: center;">21</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN3/3A</td> </tr> <tr> <td>IN11/3B</td> <td style="border: 1px solid black; text-align: center;">9</td> <td></td> <td style="border: 1px solid black; text-align: center;">20</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN2/2A</td> </tr> <tr> <td>IN10/2B</td> <td style="border: 1px solid black; text-align: center;">10</td> <td></td> <td style="border: 1px solid black; text-align: center;">19</td> <td></td> <td style="border: 1px solid black; text-align: center;">IN1/1A</td> </tr> <tr> <td>IN9/1B</td> <td style="border: 1px solid black; text-align: center;">11</td> <td></td> <td style="border: 1px solid black; text-align: center;">18</td> <td></td> <td style="border: 1px solid black; text-align: center;">ENABLE</td> </tr> <tr> <td>GND</td> <td style="border: 1px solid black; text-align: center;">12</td> <td></td> <td style="border: 1px solid black; text-align: center;">17</td> <td></td> <td style="border: 1px solid black; text-align: center;"><math>A_0</math></td> </tr> <tr> <td><math>VDD/LLS</math></td> <td style="border: 1px solid black; text-align: center;">13</td> <td></td> <td style="border: 1px solid black; text-align: center;">16</td> <td></td> <td style="border: 1px solid black; text-align: center;"><math>A_1</math></td> </tr> <tr> <td><math>A_3/SDS</math></td> <td style="border: 1px solid black; text-align: center;">14</td> <td></td> <td style="border: 1px solid black; text-align: center;">15</td> <td></td> <td style="border: 1px solid black; text-align: center;"><math>A_2</math></td> </tr> </table>	$V^+$	1		28		OUT A	OUT B	2		27		$V^-$	NC	3		26		IN8/8A	IN16/8B	4		25		IN7/7A	IN15/7B	5		24		IN6/6A	IN14/6B	6		23		IN5/5A	IN13/5B	7		22		IN4/4A	IN12/4B	8		21		IN3/3A	IN11/3B	9		20		IN2/2A	IN10/2B	10		19		IN1/1A	IN9/1B	11		18		ENABLE	GND	12		17		$A_0$	$VDD/LLS$	13		16		$A_1$	$A_3/SDS$	14		15		$A_2$	<p style="text-align: center;">INPUT BUFFER AND DECODERS      MULTIPLEXER SWITCHES</p>
$V^+$	1		28		OUT A																																																																																
OUT B	2		27		$V^-$																																																																																
NC	3		26		IN8/8A																																																																																
IN16/8B	4		25		IN7/7A																																																																																
IN15/7B	5		24		IN6/6A																																																																																
IN14/6B	6		23		IN5/5A																																																																																
IN13/5B	7		22		IN4/4A																																																																																
IN12/4B	8		21		IN3/3A																																																																																
IN11/3B	9		20		IN2/2A																																																																																
IN10/2B	10		19		IN1/1A																																																																																
IN9/1B	11		18		ENABLE																																																																																
GND	12		17		$A_0$																																																																																
$VDD/LLS$	13		16		$A_1$																																																																																
$A_3/SDS$	14		15		$A_2$																																																																																



**ABSOLUTE MAXIMUM RATINGS**

Digital Input Overvoltage:	Voltage Between Supply Pins	33V
TTL { -6V < V <sub>AH</sub> < +6V	Total Power Dissipation*	1200mW
A2 V <sub>SUPPLY</sub> (-)           -2V	Operating Temperature Ranges:	
CMOS { V <sub>SUPPLY</sub> (+)           +2V	HI-516-2	-55°C to +125°C
GND                   -2V	HI-516-5	0°C to 75°C
Analog Input Voltage:	Storage Temperature Range	-65°C to 150°C
V <sub>S</sub> { V <sub>SUPPLY</sub> (+)           +2V		
V <sub>SUPPLY</sub> (-)       -2V		

**ELECTRICAL CHARACTERISTICS** (Unless otherwise specified) Supplies = +15V, -15V; V<sub>AH</sub> (Logic Level High) = +2.4V, V<sub>AL</sub> (Logic Level Low) = +0.8V; V<sub>DD</sub>/LLS = Open (Note 6)

PARAMETER	TEMP	-55°C to +125°C			0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
V <sub>S</sub> , Analog Signal Range	Full	-15		+15	-15		+15	V
R <sub>ON</sub> , On Resistance (Note 1)	+25°C		620	750		620	750	Ω
	Full		770	1,000		700	1,000	Ω
I <sub>S</sub> (OFF), Off Input Leakage Current	+25°C		0.01			0.01		nA
	Full		0.38	50		0.38	50	nA
I <sub>D</sub> (OFF), Off Output Leakage Current	+25°C		0.035			0.035		nA
	Full		0.48	100		0.48	100	nA
I <sub>D</sub> (ON), On Channel Leakage Current	+25°C		0.04			0.04		nA
	Full		0.56	100		0.56	100	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>AL</sub> Input Low Threshold (TTL)	Full			0.8			0.8	V
V <sub>AH</sub> Input High Threshold (TTL)	Full	2.4			2.4			V
V <sub>AH</sub> Input Low Threshold (CMOS)	Full			0.3V <sub>DD</sub>			0.3V <sub>DD</sub>	V
V <sub>AH</sub> Input High Threshold (CMOS)	Full	0.7V <sub>DD</sub>			0.7V <sub>DD</sub>			V
I <sub>AH</sub> Input Leakage Current (High)	Full		0.05	1		0.05	1	μA
I <sub>AL</sub> Current (Low)	Full		4	25		4	25	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>A</sub> , Access Time	+25°C		100	150		100	150	ns
	Full		120	200		120	200	ns
t <sub>OPEN</sub> , Break before make delay	+25°C		20			20		ns
t <sub>DN</sub> (EN), Enable Delay (IN)	+25°C		100	150		100		ns
t <sub>OFF</sub> (EN), Enable Delay (OFF)	+25°C		80	125		80		ns
Settling Time (0.1%)	+25°C		250			250		ns
(0.01%)	+25°C		800			800		ns
Charge Injection (Note 2)	+25°C		0.33			0.33		pC
Off Isolation (Note 3)	+25°C		90			90		dB
C <sub>S</sub> (OFF), Channel Input Capacitance	+25°C		2.5			2.5		pF
C <sub>D</sub> (OFF), Channel Output Capacitance	+25°C		18			18		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		5			5		pF
C <sub>DS</sub> (OFF), Input to Output Capacitance	+25°C		0.02			0.02		pF
<b>POWER REQUIREMENTS</b>								
P <sub>D</sub> , Power Dissipation	Full		525			525		mW
I <sup>+</sup> , Current (Note 4)	Full		17.5	25		17.5	30	mA
I <sup>-</sup> , Current (Note 4)	Full		17.5	25		17.5	30	mA
I <sup>+</sup> , Standby (Note 5)	Full		17.0	25		17.0	30	mA
I <sup>-</sup> , Standby (Note 5)	Full		17.0	25		17.0	30	mA

NOTES:

- V<sub>IN</sub> = ± 10V, I<sub>OUT</sub> = -100 μA
- V<sub>IN</sub> = 0V, C<sub>L</sub> = 100pF, Enable input pulse = 3V, f = 500kHz
- V<sub>EN</sub> = 0.8V, V<sub>S</sub> = 3V<sub>RMS</sub>, f = 500kHz, C<sub>L</sub> = 40pF, R<sub>L</sub> = 1k, Pin 3 grounded
- V<sub>EN</sub> = +2.4V
- V<sub>EN</sub> = 0.8V
- V<sub>DD</sub>/LLS Pin = Open or Grounded for TTL Compatibility  
V<sub>DD</sub>/LLS Pin = V<sub>DD</sub> for CMOS Compatibility





HI-516 USED AS A 16-CHANNEL MULTIPLEXER OR  
8 CHANNEL DIFFERENTIAL MULTIPLEXER \*

USE A <sub>3</sub> AS DIGITAL ADDRESS INPUT					ON CHANNEL TO	
ENABLE	A <sub>3</sub>	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	OUT A	OUT B
L	X	X	X	X	NONE	NONE
H	L	L	L	L	1A	NONE
H	L	L	L	H	2A	NONE
H	L	L	H	L	3A	NONE
H	L	L	H	H	4A	NONE
H	L	H	L	L	5A	NONE
H	L	H	L	H	6A	NONE
H	L	H	H	L	7A	NONE
H	L	H	H	H	8A	NONE
H	H	L	L	L	NONE	1B
H	H	L	L	H	NONE	2B
H	H	L	H	L	NONE	3B
H	H	L	H	H	NONE	4B
H	H	H	L	L	NONE	5B
H	H	H	L	H	NONE	6B
H	H	H	H	L	NONE	7B
H	H	H	H	H	NONE	8B

\* For 16-Channel single-ended function, tie 'out A' to 'out B', for dual 8-channel function use the A<sub>3</sub> address pin to select between MUX A and MUX B, where MUX A is selected with A<sub>3</sub> low.

HI-516 USED AS A DIFFERENTIAL  
8-CHANNEL MULTIPLEXER

A <sub>3</sub> CONNECT TO V <sup>-</sup> SUPPLY				ON CHANNEL TO	
ENABLE	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	OUT A	OUT B
L	X	X	X	NONE	NONE
H	L	L	L	1A	1B
H	L	L	H	2A	2B
H	L	H	L	3A	3B
H	L	H	H	4A	4B
H	H	L	L	5A	5B
H	H	L	H	6A	6B
H	H	H	L	7A	7B
H	H	H	H	8A	8B



# HI-518

## 8 Channel/Differential 4 Channel CMOS High Speed Analog Multiplexer

FEATURES	DESCRIPTION												
<ul style="list-style-type: none"> <li>● ACCESS TIME (TYP) 80ns</li> <li>● SETTLING TIME (0.1%) 250ns</li> <li>● LOW LEAKAGE <math>I_S</math> (OFF) 50pA</li> <li>    <math>I_D</math> (OFF) 100pA</li> <li>● LOW CAPACITANCE (TYP) <math>C_S</math> (OFF) 2pF</li> <li>    <math>C_D</math> (OFF) 10pF</li> <li>● HIGH OFF ISOLATION @ (1MHz) 75dB</li> <li>● SINGLE ENDED TO DIFFERENTIAL MODE SELECTABLE (SDS)</li> <li>● LOGIC LEVEL SELECTABLE (LLS)</li> <li>● LOW CHARGE INJECTION 0.3pC</li> </ul>	<p>The HI-518 is a monolithic, high performance, high speed Analog Multiplexer, constructed utilizing the Harris Dielectrically isolated CMOS process.</p> <p>This device has the added feature that it can be user programmed either as a single ended 8-channel multiplexer by connecting 'out A' to 'out B' and using A2 as a digital address input, or as a 4-channel differential multiplexer by connecting A2 to the V<sup>-</sup> supply.</p> <p>TTL or CMOS compatibility is also selectable. Low leakage current, <math>I_D</math> off &lt; 100pA @ 25°C, and fast settling, 250ns to 0.1%, characteristics of this device make it an ideal choice for high speed data acquisition systems, precision instrumentation and industrial process controls.</p> <p>The HI-518 is available in an 18 lead Dual-in-Line Package. The HI-518-5 is specified for operation over 0°C to +75°C, and the HI-518-2 over -55°C to +125°C. Processing to MIL-STD-883A Class B screening is available by selecting the HI-518-8.</p>												
APPLICATIONS													
<ul style="list-style-type: none"> <li>● DATA ACQUISITION SYSTEMS</li> <li>● INDUSTRIAL CONTROLS</li> <li>● TELEMETRY</li> </ul>													
PINOUT	FUNCTIONAL DIAGRAM												
<p style="text-align: center;">TOP VIEW</p>	<table border="1" style="margin-left: 20px;"> <caption>A2 DECODE</caption> <tr><td>A2</td><td>Q</td><td>Q̄</td></tr> <tr><td>H</td><td>H</td><td>L</td></tr> <tr><td>L</td><td>L</td><td>H</td></tr> <tr><td>V<sup>-</sup></td><td>L</td><td>L</td></tr> </table>	A2	Q	Q̄	H	H	L	L	L	H	V <sup>-</sup>	L	L
A2	Q	Q̄											
H	H	L											
L	L	H											
V <sup>-</sup>	L	L											



**ABSOLUTE MAXIMUM RATINGS**

Digital Input Overvoltage:		Voltage Between Supply Pins	33V
TTL	-6V < V <sub>AH</sub> < +6V	Total Power Dissipation*	725mW
	A2 V <sub>SUPPLY</sub> (-)	-2V	
CMOS	V <sub>SUPPLY</sub> (+)	+2V	Operating Temperature Ranges:
	GND	-2V	HI-518-2 -55°C to +125°C
Analog Input Voltage:		HI-518-5 0°C to 75°C	
V <sub>S</sub>	V <sub>SUPPLY</sub> (+)	+2V	Storage Temperature Range -65°C to 150°C
	V <sub>SUPPLY</sub> (-)	-2V	*Derate 13.6mW/°C above T <sub>A</sub> = 120°C

**ELECTRICAL CHARACTERISTICS**

(Unless otherwise specified) Supplies = +15V, -15V; V<sub>AH</sub> (Logic Level High) = +2.4V, V<sub>AL</sub> (Logic Level Low) = +0.8V; V<sub>DD</sub>/LLS = Open (Note 6).

PARAMETER	TEMP	-55°C to +125°C			0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>								
V <sub>S</sub> Analog Signal Range	Full	-15		+15	-15		+15	V
R <sub>ON</sub> On Resistance (Note 1)	+25°C		480	750		480	750	Ω
	Full		700	1000		700	1000	Ω
I <sub>S</sub> (OFF) Off Input Leakage Current	+25°C		0.05			0.05		nA
	Full		0.60	50		0.60	50	nA
I <sub>D</sub> (OFF) Off Output Leakage Current	+25°C		0.10			0.10		nA
	Full		0.30	50		0.30	50	nA
I <sub>D</sub> (ON) On Channel Leakage Current	+25°C		0.10			0.10		nA
	Full		0.30	50		0.30	50	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>								
V <sub>AL</sub> Input Low Threshold (TTL)	Full			0.8			0.8	V
V <sub>AH</sub> Input High Threshold (TTL)	Full	2.4			2.4			V
V <sub>AL</sub> Input Low Threshold (CMOS)	Full			0.3V <sub>DD</sub>			0.3V <sub>DD</sub>	V
V <sub>AH</sub> Input High Threshold (CMOS)	Full	0.7V <sub>DD</sub>			0.7V <sub>DD</sub>			V
I <sub>AH</sub> Input Leakage Current (High)	Full		0.05	1		0.05	1	μA
I <sub>AH</sub> Input Leakage Current (Low)	Full		4	20		4	20	μA
<b>SWITCHING CHARACTERISTICS</b>								
t <sub>A</sub> , Access Time	+25°C		80	125		80	125	ns
	Full		110	150		110	150	ns
t <sub>OPEN</sub> , Break before make Delay	+25°C		20			20		ns
t <sub>ON</sub> (EN), Enable Delay (ON)	+25°C		80	150		80	150	ns
t <sub>OFF</sub> (EN), Enable Delay (OFF)	+25°C		60	125		60	125	ns
Settling Time (0.1%)	+25°C		250			250		ns
(0.01%)	+25°C		800			800		ns
Charge Injection (Note 2)	+25°C		0.3			0.3		pC
Off Isolation (Note 3)	+25°C		86			86		dB
C <sub>S</sub> (OFF) Channel Input Capacitance	+25°C		1.9			1.9		pF
C <sub>D</sub> (OFF) Channel Output Capacitance	+25°C		10			10		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		3			3		pF
C <sub>DS</sub> (OFF) Input to Output Capacitance	+25°C		0.02			0.02		pF
<b>POWER REQUIREMENTS</b>								
P <sub>D</sub> , Power Dissipation	Full		360	450		360	540	mW
I <sub>+</sub> , Current (Note 4)	Full		12	15		12	18	mA
I <sub>-</sub> , Current (Note 4)	Full		12	15		12	18	mA
I <sub>+</sub> , Standby (Note 5)	Full		11.5	15		11.5	18	mA
I <sub>-</sub> , Standby (Note 5)	Full		11.5	15		11.5	18	mA

**NOTES:**

- V<sub>IN</sub> = ±10V, I<sub>OUT</sub> = -100μA
- V<sub>IN</sub> = 0V, C<sub>L</sub> = 100pF, Enable Input pulse = 3V, f = 500kHz.
- V<sub>EN</sub> = 0.8V, V<sub>S</sub> = 3VRMS, f = 500kHz, C<sub>L</sub> = 40pF, R<sub>L</sub> = 1k. Due to the pin to pin capacitance between IN 8/4B (Pin 3) and Out B (Pin 2) channel 8/4B exhibits 60dB of Off Isolation under the above test conditions.
- V<sub>EN</sub> = +2.4V.
- V<sub>EN</sub> = 0.8V.
- V<sub>DD</sub>/LLS Pin = Open or grounded for TTL compatibility. V<sub>DD</sub>/LLS Pin = V<sub>DD</sub> for CMOS compatibility.

HI-518 USED AS 8 CHANNEL MULTIPLEXER OR  
4 CHANNEL DIFFERENTIAL MULTIPLEXER

USE A <sub>2</sub> AS DIGITAL ADDRESS INPUT				ON CHANNEL TO	
ENABLE	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	OUT A	OUT B
L	X	X	X	NONE	NONE
H	L	L	L	1A	NONE
H	L	L	H	2A	NONE
H	L	H	L	3A	NONE
H	L	H	H	4A	NONE
H	H	L	L	NONE	1B
H	H	L	H	NONE	2B
H	H	H	L	NONE	3B
H	H	H	H	NONE	4B

HI-518 USED AS DIFFERENTIAL  
4 CHANNEL MULTIPLEXER

A <sub>2</sub> CONNECT TO V <sup>-</sup> SUPPLY			ON CHANNEL TO	
ENABLE	A <sub>1</sub>	A <sub>0</sub>	OUT A	OUT B
L	X	X	NONE	NONE
H	L	L	1A	1B
H	L	H	2A	2B
H	H	L	3A	3B
H	H	H	4A	4B

FEATURES	
• CROSSTALK (10MHz)	> 60dB
• FAST ACCESS TIME	150ns
• FAST SETTLING TIME (0.01%)	600ns
• TTL COMPATIBLE	

**DESCRIPTION**

The HI-524 is a four channel CMOS analog multiplexer designed to process single-ended video signals with bandwidths up to 10MHz. The chip includes a 1 of 4 decoder for channel selection and an Enable input to inhibit all channels (chip select).

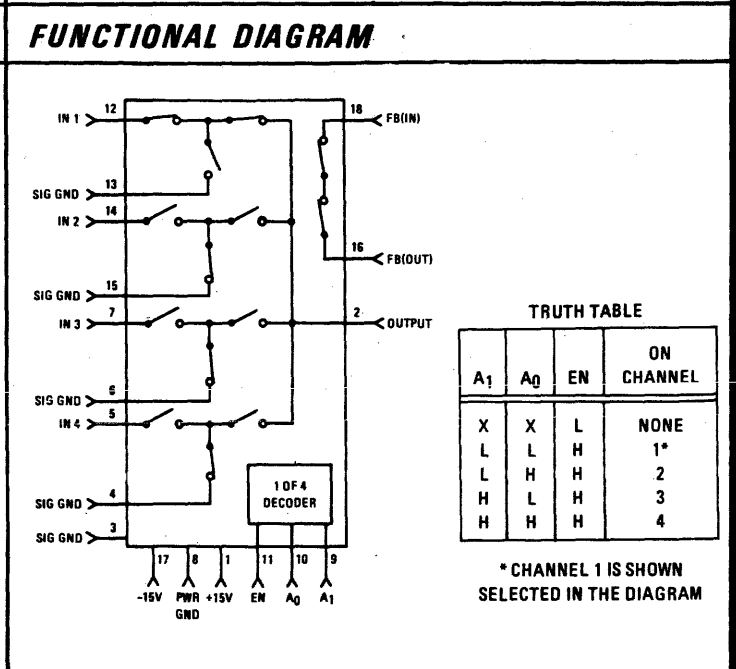
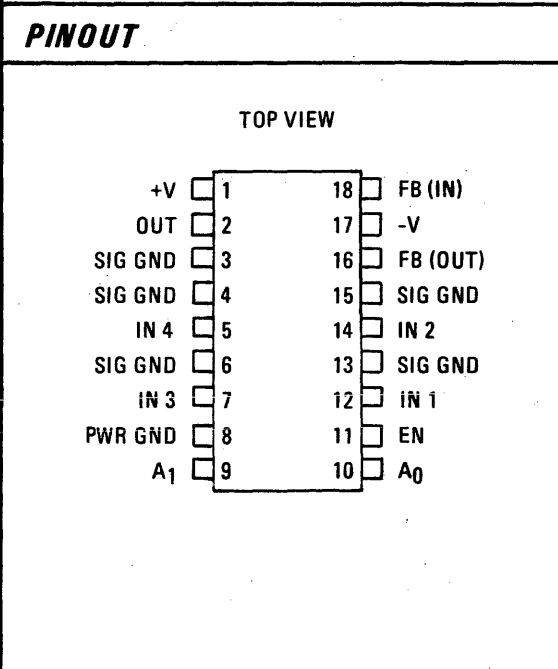
Three CMOS transmission gates are used in each channel, as compared to the single gate in more conventional CMOS multiplexers. This provides a double barrier to the unwanted coupling of signals from each input to the output. In addition, Dielectric Isolation (DI) processing helps to insure that Crosstalk exceeds 60dB at 10MHz.

The HI-524 is designed to operate into a wideband buffer amplifier such as the HARRIS HA-5190. The multiplexer chip includes two "on" switches in series, for use as a feedback element with the amplifier. This feedback resistance matches and tracks the channel RON resistance, to minimize the amplifier VOS and its variation with temperature.

The HI-524 is well suited to the rapid switching of video signals in telemetry, instrumentation, radar and video systems. It is packaged in an 18 pin ceramic DIP and operates on ±15V supplies.

The performance levels available are: HI1-524-2, -55°C to +125°C operating range; HI1-524-5, 0°C to +75°C operating range and HI1-524-8, -55°C to +125°C operating range plus 100% screening per MIL-STD-883/Method 5004/Class B. Chips for hybrid applications are designated HI0-524-6.

- APPLICATIONS**
- WIDEBAND SWITCHING
- RADAR
  - TV VIDEO
  - ECM



**SPECIFICATIONS**



**ABSOLUTE MAXIMUM RATINGS**

Digital Input Overvoltage: -6V < V <sub>AH</sub> < +6V	Either Supply to Ground	16.5V
Analog Input (V <sub>S</sub> ) or Output (V <sub>O</sub> ) +V <sub>SUPPLY</sub> +2V -V <sub>SUPPLY</sub> -2V	Total Power Dissipation*	750mW
Voltage Between Supply Pins	Operating Temperature Range: HI-524-2, -8 HI-524-5	-55°C to +125°C 0°C to 75°C
	Storage Temperature Range	-65°C to 150°C
	*Derate 13.6mW/°C above T <sub>A</sub> = 120°C	

**ELECTRICAL CHARACTERISTICS** (Unless otherwise specified) Supplies = +15V, -15V; V<sub>AH</sub> (Logic Level High) = +2.4V, V<sub>AL</sub> = (Logic Level Low) = +0.8V; V<sub>EN</sub> = +2.4V

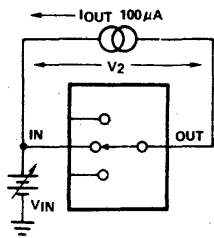
PARAMETER	TEMP	HI-524-2, -8 -55°C to +125°C			HI-524-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<u>Analog Channel Characteristics</u>								
V <sub>S</sub> , Analog Signal Range	Full	-10		+10	-10		+10	V
RON, On Resistance (Note 1)	+25°C		700			700		Ω
I <sub>S</sub> (OFF), Off Input Leakage Current (Note 2)	Full		0.2	1.5K		0.2	1.5K	Ω
I <sub>D</sub> (OFF), Off Output Leakage Current (Note 2)	+25°C		0.2	50		0.2	50	nA
I <sub>D</sub> (ON), On Channel Leakage Current (Note 2)	Full		0.2	50		0.2	50	nA
I <sub>D</sub> (ON), On Channel Leakage Current (Note 2)	+25°C		0.7	50		0.7	50	nA
I <sub>D</sub> (ON), On Channel Leakage Current (Note 2)	Full		0.7	50		0.7	50	nA
3dB Bandwidth: (Note 3)	Full		20			20		MHz
<u>Digital Input Characteristics</u>								
V <sub>AL</sub> Input Low Threshold (TTL)	Full			0.8			0.8	V
V <sub>AH</sub> Input High Threshold (TTL)	Full	2.4			2.4			V
I <sub>AH</sub> Input Leakage Current (High)	Full		0.05	1		0.05	1	μA
I <sub>AL</sub> Current (Low)	Full		4	25		4	25	μA
<u>Switching Characteristics</u>								
t <sub>A</sub> , Access Time (Note 4)	+25°C		150	300		150	300	ns
t <sub>OPEN</sub> , Break before make delay (Note 4)	Full							ns
t <sub>ON</sub> (EN), Enable Delay (ON), R <sub>L</sub> = 500Ω	+25°C		20			20		ns
t <sub>OFF</sub> (EN), Enable Delay (OFF), R <sub>L</sub> = 500Ω	+25°C		180	300		180		ns
Settling Time (0.1%) (Note 4)	+25°C		180	250		180		ns
Settling Time (0.01%) (Note 4)	+25°C		200			200		ns
Crosstalk (Note 5)	+25°C		600			600		ns
CS (OFF), Channel Input Capacitance	+25°C		-65			-65		dB
CD (OFF), Channel Output Capacitance	+25°C			6			6	pF
CA, Digital Input Capacitance	+25°C			4			4	pF
	+25°C			5			5	pF
<u>Power Requirements</u>								
PD, Power Dissipation	Full		540			540		mW
I <sup>+</sup> , Current (V <sub>EN</sub> = 2.4V) (Note 6)	Full		18	25		18	25	mA
I <sup>-</sup> , Current (V <sub>EN</sub> = 2.4V) (Note 6)	Full		18	25		18	25	mA
I <sup>+</sup> , Standby (V <sub>EN</sub> = 0.8V) (Note 6)	Full		18	25		18	25	mA
I <sup>-</sup> , Standby (V <sub>EN</sub> = 0.8V) (Note 6)	Full		18	25		18	25	mA

- V<sub>IN</sub> = 0V; I<sub>OUT</sub> = 100μA (See Test Circuit # 1)
- V<sub>O</sub> = ±10V; V<sub>S</sub> = ±10V (See Test Circuits # 2, 3, 4)
- MUX output is buffered with HA-5190 as shown in Applications section.
- (See Test Circuit # 5)
- V<sub>IN</sub> = 10MHz, 3V<sub>p-p</sub> on one channel, with any other channel selected. (Worst case is channel 3 selected with input on channel 4.) MUX output is buffered with HA-5190 as shown in Applications section. Terminate all channels with 75Ω.
- Supply currents vary less than 0.5mA for switching rates from DC to 2MHz.

(UNLESS OTHERWISE SPECIFIED  $T_A = 25^\circ\text{C}$ ,  $V_{\text{SUPPLY}} = \pm 15\text{V}$ ,  $V_{\text{AH}} = 2.4\text{V}$ ,  $V_{\text{AL}} = 0.8\text{V}$ )

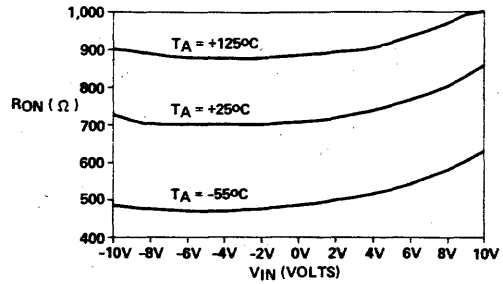
ON RESISTANCE

TEST CIRCUIT NO. 1

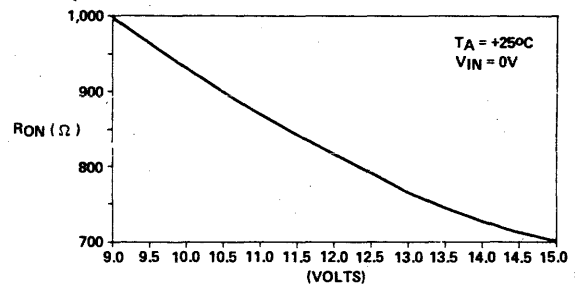


$$R_{\text{ON}} = \frac{V_2}{100\mu\text{A}}$$

ON RESISTANCE VS. ANALOG INPUT VOLTAGE

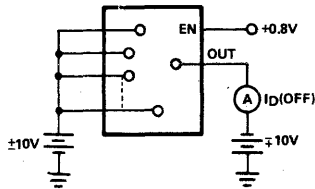


ON RESISTANCE VS. SUPPLY VOLTAGE



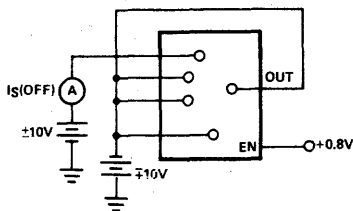
LEAKAGE CURRENT

TEST CIRCUIT NO. 2\*

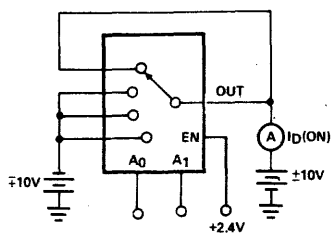


\*Two measurements per channel:  
+10V/-10V and -10V/+10V.  
(Two measurements per device for ID(OFF):  
+10V/-10V and -10V/+10V.)

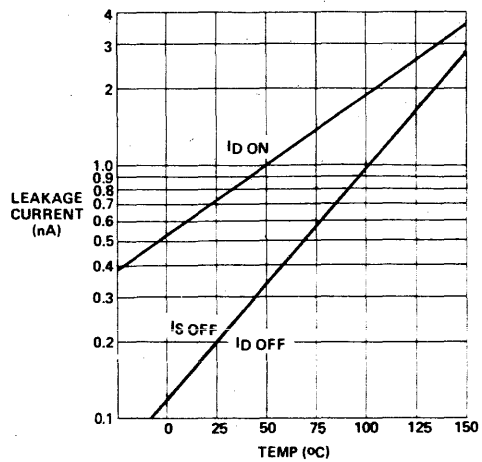
TEST CIRCUIT NO. 3\*



TEST CIRCUIT NO. 4\*

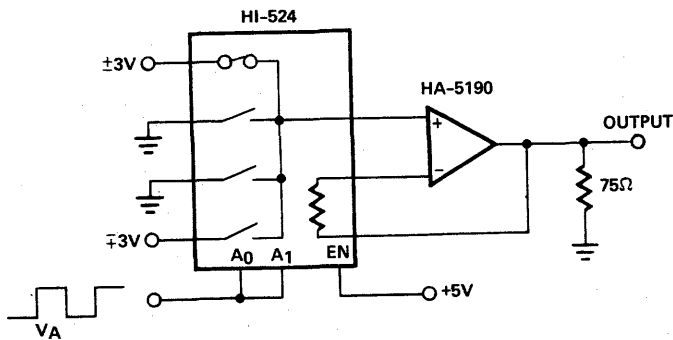


LEAKAGE CURRENT VS. TEMPERATURE

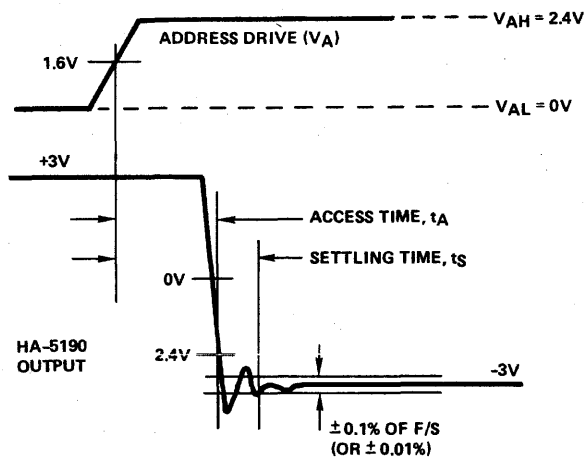




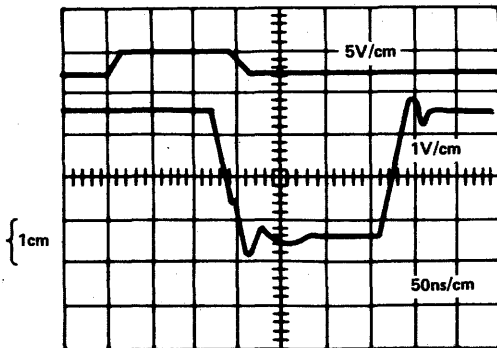
TEST CIRCUIT NO. 5  
 SETTLING TIME  
 ACCESS TIME  
 BREAK-BEFORE-MAKE DELAY



(USE DIFFERENTIAL COMPARATOR PLUG-IN ON SCOPE FOR SETTLING TIME MEASUREMENT.)

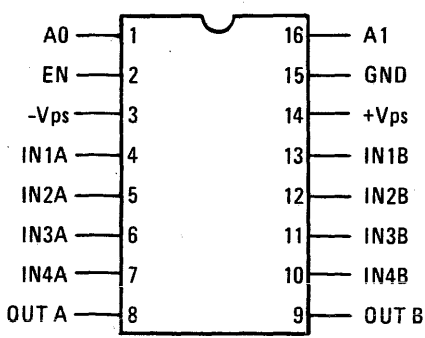
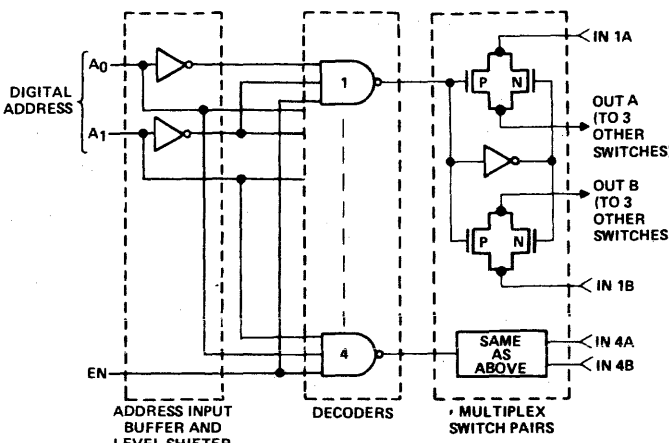


ACCESS TIME





## Monolithic, Four Channel, Low Level, Differential Multiplexer

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• DIFFERENTIAL PERFORMANCE, TYP.:                             <ul style="list-style-type: none"> <li>• LOW <math>\Delta R_{ON}</math>, +125°C <span style="float: right;">5.5<math>\Omega</math></span></li> <li>• LOW <math>\Delta I_{D(ON)}</math>, +125°C <span style="float: right;">0.6nA</span></li> <li>• LOW <math>\Delta</math>(CHARGE INJECTION) <span style="float: right;">0.1pC</span></li> <li>• LOW CROSSTALK <span style="float: right;">-120dB</span></li> </ul> </li> <li>• SETTLING TIME, <math>\pm 0.01\%</math> <span style="float: right;">900ns</span></li> <li>• WIDE SUPPLY RANGE <span style="float: right;"><math>\pm 5V</math> TO <math>\pm 18V</math></span></li> <li>• BREAK-BEFORE-MAKE SWITCHING</li> <li>• NO LATCH-UP</li> </ul>	<p>The Harris HI-539 is a monolithic, four channel, differential multiplexer. Two digital inputs are provided for channel selection, plus an Enable input to disconnect all channels.</p> <p>Performance is guaranteed for each channel over the range <math>\pm 10V</math>, but is optimized for low level differential signals. Leakage current, for example, which varies slightly with input voltage, has its distribution centered at zero for zero input volts.</p> <p>In most monolithic multiplexers, the net differential offset due to thermal effects becomes significant for low level signals. This problem is minimized in the HI-539 by symmetrical placement of critical circuitry with respect to the few heat producing devices.</p> <p>The HI-539 will be offered in both commercial and military temperature ranges, with screening available for MIL-STD-883, Class B. Supply voltages are <math>\pm 15V</math> and power consumption is only 2.5mW. The package is a 16 pin ceramic DIP.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• LOW LEVEL DATA ACQUISITION</li> <li>• PRECISION INSTRUMENTATION</li> <li>• TEST SYSTEMS</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;">TOP VIEW</p> 	

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**ABSOLUTE MAXIMUM RATINGS**

Voltage Between Supply Pins ( $V_{PS+}, V_{PS-}$ )	40V	Internal Power Dissipation*	725mW
Voltage from either Supply to Ground	20V	Operating Temperature Range	
Analog Input Voltage, $V_S$	$V_{PS-} \leq V_S \leq V_{PS+}$	HI-539-2, -8	-55°C to +125°C
Digital Input Voltage, $V_A$	$V_{PS-} \leq V_A \leq V_{PS+}$	HI-539-4	-25°C to +85°C
		HI-539-5	0°C to +75°C
Storage Temperature Range	-65°C to +150°C	*Derate 9.6mW/°C above $T_A = 95°C$	

**ELECTRICAL CHARACTERISTICS**

(Unless otherwise specified) Supplies =  $\pm 15V$ ,  $V_{EN} = +4.0V$ ,  $V_{AH}$  (Logic Level High) = +4.0V,  $V_{AL}$  (Logic Level Low) = +0.8V. See the Performance Characteristics Section for test circuits and conditions. Selected parameters are defined in the Definitions Section.

PARAMETER	TEMP	HI-539-2, -8		HI-539-4, -5		UNITS	
		TYP	MAX (MIN)	TYP	MAX (MIN)		
<b>ANALOG CHANNEL CHARA'S</b>							
$V_S$ , Analog Signal Range	Full		(-10)/+10		(-10)/+10	V	
$R_{ON}$ , On Resistance	$V_{IN} = 0V$	+25°C	650	850	650	850	$\Omega$
	$V_{IN} = \pm 10V$	+25°C	700	900	700	900	$\Omega$
	$V_{IN} = 0V$	Full	950	1.3K	800	1K	$\Omega$
	$V_{IN} = \pm 10V$	Full	1.1k	1.4k	900	1.1k	$\Omega$
$\Delta R_{ON}$ [Side A - Side B]	$V_{IN} = 0V$	+25°C	4.0	24	4.0	24	$\Omega$
	$V_{IN} = \pm 10V$	+25°C	4.5	27	4.5	27	$\Omega$
	$V_{IN} = 0V$	Full	4.75	28	4.0	24	$\Omega$
	$V_{IN} = \pm 10V$	Full	5.5	33	4.5	27	$\Omega$
$I_{S(OFF)}$ , Off Input Leakage Current (Note 1)	Condition 0V	+25°C	30	200	30	200	pA
	Condition $\pm 10V$	+25°C	100		100		pA
	Condition 0V	Full	2	10	0.2	1	nA
	Condition $\pm 10V$	Full	5	25	0.5	2.5	nA
$\Delta I_{S(OFF)}$ , [Side A - Side B]	Condition 0V	+25°C	3	100	3	100	pA
	Condition $\pm 10V$	+25°C	10		10		pA
	Condition 0V	Full	0.2	2	0.02	0.2	nA
	Condition $\pm 10V$	Full	0.5	5	0.05	0.5	nA
$I_{D(OFF)}$ , Off Output Leakage Current (Note 1)	Condition 0V	+25°C	30	200	30	200	pA
	Condition $\pm 10V$	+25°C	100		100		pA
	Condition 0V	Full	2	10	0.2	1	nA
	Condition $\pm 10V$	Full	5	25	0.5	2.5	nA
$\Delta I_{D(OFF)}$ , [Side A - Side B]	Condition 0V	+25°C	3	100	3	100	pA
	Condition $\pm 10V$	+25°C	10		10		pA
	Condition 0V	Full	0.2	2	0.02	0.2	nA
	Condition $\pm 10V$	Full	0.5	5	0.05	0.5	nA
$I_{D(ON)}$ , On Channel Leakage Current (Note 1)	Condition 0V	+25°C	50	200	50	200	pA
	Condition $\pm 10V$	+25°C	150		150		pA
	Condition 0V	Full	5	25	0.5	2.5	nA
	Condition $\pm 10V$	Full	6	40	0.8	4.0	nA
$\Delta I_{D(ON)}$ [Side A - Side B]	Condition 0V	+25°C	10	100	10	100	pA
	Condition $\pm 10V$	+25°C	30		30		pA
	Condition 0V	Full	0.5	5	0.05	0.5	nA
	Condition $\pm 10V$	Full	0.6	6	0.08	0.8	nA
$\Delta V_{OS}$ , Differential Offset Voltage	+25°C	0.02	0.04	0.02	0.04	$\mu V$	
	Full	0.70	10	0.08	1.0	$\mu V$	



PARAMETER	TEMP	HI-539-2, -8		HI-539-4, -5		UNITS
		TYP	MAX (MIN)	TYP	MAX (MIN)	
<b>DIGITAL INPUT CHARACTERISTICS</b>						
V <sub>AL</sub> , Input Low Threshold	Full		0.8		0.8	V
V <sub>AH</sub> , Input High Threshold	Full		(4.0)		(4.0)	V
I <sub>AH</sub> , Input Leakage Current (High)	Full		1		1	μA
I <sub>AL</sub> , Input Leakage Current (Low)	Full		1		1	μA
<b>SWITCHING CHARACTERISTICS</b>						
T <sub>A</sub> , Access Time	+25°C	250	750	250	750	ns
	Full	450	1,000	450	1,000	ns
T <sub>open</sub> , Break-Before-Make Delay	+25°C	85	(30)	85	(30)	ns
	Full		(30)		(30)	ns
T <sub>ON(EN)</sub> , Enable Delay On	+25°C	250	750	250	750	ns
	Full		1,000		1,000	ns
T <sub>OFF(EN)</sub> , Enable Delay Off	+25°C	160	650	160	650	ns
	Full		900		900	ns
Settling Time, to ± 0.01%	+25°C	0.9		0.9		μs
Charge Injection (Output)	Full	3		3		pC
Δ Charge Injection (Output)	Full	0.1		0.1		pC
Charge Injection (Input)	Full	10		10		pC
Differential Crosstalk (Note 3)	+25°C	124		124		dB
Single Ended Crosstalk (Note 3)	+25°C	105		100		dB
C <sub>S(OFF)</sub> , Channel Input Capacitance	Full	5		5		pF
C <sub>D(OFF)</sub> , Channel Output Capacitance	Full	7		7		pF
C <sub>D(ON)</sub> , Channel On Output Capacitance	Full	17		17		pF
C <sub>D(S)</sub> , Input to Output Capacitance (Note 4)	Full	0.08		0.08		pF
C <sub>A</sub> , Digital Input Capacitance	Full	3		3		pF
<b>POWER REQUIREMENTS</b>						
P <sub>D</sub> , Power Dissipation	+25°C	2.5		2.5		mW
	Full		45		45	mW
I <sub>+</sub> Current	+25°C	0.150		0.150		mA
	Full		2.0		2.0	mA
I <sub>-</sub> Current	+25°C	0.001		0.001		mA
	Full		1.0		1.0	mA
± V, Supply Voltage Range	Full	± 15	(± 5)/ ± 18	± 15	(± 5)/ ± 18	V

NOTES

1. See Test Circuits #2, 3, 4. The condition ± 10V means:

I<sub>S(OFF)</sub> and I<sub>D(OFF)</sub>: (V<sub>S</sub> = +10V, V<sub>D</sub> = -10V), then  
 (V<sub>S</sub> = -10V, V<sub>D</sub> = +10V)  
 I<sub>D(ON)</sub>: (+10V, then -10V)

2. ΔV<sub>QS</sub> (Exclusive of thermocouple effects) =  
 R<sub>ON</sub> ΔI<sub>D(ON)</sub> + I<sub>D(ON)</sub> ΔR<sub>ON</sub>.

See Applications section for discussion of additional V<sub>QS</sub> error.

3. V<sub>IN</sub> = 1kHz, 15V<sub>p-p</sub> on all but the selected channel. See Test Circuit #9.

4. Calculated from typical Single-Ended Crosstalk performance.



# HI-562A

## 12 Bit High Speed Monolithic Digital-to-Analog Converter

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• OUTPUT CURRENT 2mA, F.S.</li> <li>• MONOLITHIC CONSTRUCTION</li> <li>• EXTREMELY FAST SETTLING 300ns TO 0.01% (TYP.)</li> <li>• LOW GAIN DRIFT <math>\pm 10\text{ppm}/^\circ\text{C}</math> (MAX.)</li> <li>• EXCELLENT LINEARITY <math>\pm 1/2</math> LSB (MAX.)</li> <li>• DESIGNED FOR MINIMUM GLITCHES</li> <li>• MONOTONIC OVER TEMPERATURE</li> <li>• NOTE: HI-562A IS RECOMMENDED FOR NEW DESIGNS</li> </ul>	<p>The Harris HI-562A is the first monolithic digital-to-analog converter to combine both ultra-high speed performance and 12-bit accuracy on the same chip. The HI-562A's fast output current settling of 300ns to 0.01% is achieved using dielectric isolation processing to reduce internal parasitics for fast rise and fall times during switching. Output glitches are minimized in the HI-562A by incorporating equally weighted current sources switched into an R-2R ladder network for symmetrical turn-ON and turn-OFF switching times. This creates within the chip a very uniform constant thermal distribution for excellent linearity and also completely eliminates thermal transients during switching. High stability thin film resistor processing together with laser trimming provide the HI-562A with guaranteed true 12-bit linearity to within <math>\pm 1/2</math> LSB maximum at <math>+25^\circ\text{C}</math> for -4 and -5 parts, and to within <math>\pm 1/4</math> LSB maximum at <math>+25^\circ\text{C}</math> for -2 and -8 parts. The HI-562A is recommended as a replacement for higher cost hybrid and modular units for increased reliability and accuracy in applications such as CRT displays, precision instruments and data acquisition systems requiring throughput rates as high as 3.3 MHz for full range transitions. Its small size makes it an ideal choice as the heart of high speed A/D converter designs or as a building block in high speed or high resolution industrial process control systems. The HI-562A is also ideally suited for aircraft and space instrumentation where operation over a wide temperature range is required.</p> <p>The HI-562A-5 is specified for operation over <math>0^\circ\text{C}</math> to <math>+75^\circ\text{C}</math>, the HI-562A-4 over <math>-25^\circ\text{C}</math> to <math>+85^\circ\text{C}</math> and the HI-562A-2 and HI-562A-8 over <math>-55^\circ\text{C}</math> to <math>+125^\circ\text{C}</math>. Processing MIL-STD-883A Class B screening is available by selecting the HI-562A-8. All are available in a hermetically sealed 24-lead dual-in-line package.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• CRT DISPLAY GENERATION</li> <li>• HIGH SPEED A/D CONVERTERS</li> <li>• VIDEO SIGNAL RECONSTRUCTION</li> <li>• WAVEFORM SYNTHESIZERS</li> <li>• HIGH SPEED DATA ACQUISITION</li> <li>• HIGH-REL APPLICATIONS</li> <li>• PRECISION INSTRUMENTS</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;">Section 11 for Packaging</p> <p style="text-align: center;">TOP VIEW</p> <p style="text-align: center;"><b>HI-562A</b></p> <p>*Pin 3 connected to bottom case for high frequency shielding.</p>	

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# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Referred to Ground)<sup>1</sup>

Power Supply Inputs	V <sub>ps+</sub>	+20V	Power Dissipation	P <sub>d</sub> , Package	1000mW
	V <sub>ps-</sub>	-20V	Operating Temperature Range		
Reference Inputs	V <sub>REF</sub> (Hi)	±V <sub>ps</sub>	HI-562A-2		-55°C to +125°C
Digital Inputs	Bits 1-12	-1V, +12V	HI-562A-4		-25°C to +85°C
	CMOS/TTL Logic Select	-1V, +12V	HI-562A-5		0°C to +75°C
Outputs	Pins 7, 8, 10, 11	±V <sub>ps</sub>	HI-562A-8		-55°C to +125°C
	Pin 9	+V <sub>ps</sub> , -5V	Storage Temperature Range		-65°C to +150°C

## ELECTRICAL CHARACTERISTICS (@ +25°C, V<sub>ps+</sub> = +5V, V<sub>ps-</sub> = -15V, V<sub>REF</sub> = +10V, pin 2 tied to pin 12 unless otherwise noted)

PARAMETER	CONDITIONS	HI-562A-2/HI-562A-8			HI-562A-4/HI-562A-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

### INPUT CHARACTERISTICS

Digital Inputs (3)	Bit ON "Logic 1" Bit OFF "Logic 0"							
TTL	Input Voltage (2) Logic "1" Logic "0"	2.0		0.8	2.0		0.8	V V
	Input Current (2) Logic "1" Logic "0"		Pin 2 tied to Pin 12	20 -50		±500 -100	20 -50	±500 -100
CMOS	Input Voltage Logic "1" Logic "0"	0.7V <sub>ps+</sub>		0.3V <sub>ps+</sub>	0.7V <sub>ps+</sub>		0.3V <sub>ps+</sub>	V V
	Input Current Logic "1" Logic "0"		Connect pin 2 to pin 1 for V <sub>ps+</sub> ≥ 9.5V. Otherwise (for CMOS levels below 8V), connect pin 2 to pin 12.	20 -50		±500 -100	20 -50	±500 -100
Reference Input	Input Resistance		19.95K			19.95K		Ω
	Input Voltage		+10			+10		V

### TRANSFER CHARACTERISTICS

Resolution	Over full temp. range			12			12	Bits
Nonlinearity (3)	@ +25°C			±1/4		±1/4	±1/2	LSB
	Over full temp. range		±1/2	±1			±1	
Differential Nonlinearity (3)	@ +25°C			±1/4		±1/4	±1/2	LSB
	Over full temp. range			MONOTONICITY GUARANTEED				
Relative Accuracy (6)	With 50Ω (1%) Trim Resistors							
	Gain Error		±0.24	±0.25		±0.24	±0.25	% FSR (4)
	Bipolar Offset Error		±0.24	±0.25		±0.24	±0.25	
Unipolar Offset Error	All Bits OFF		±0.12	±0.05		±0.12	±0.05	
Adjustment Range	See Operating Instructions							
	Gain		±0.25			±0.25		% FSR
	Bipolar Offset		±0.5			±0.5		
Temperature Stability	Drift specified with internal span resistors for voltage output							
	Gain Drift (3)		±6	±10			±10	ppm of FSR/°C
	Offset Drift (3)							
	Unipolar Offset			±2			±2	
Bipolar Offset	All Bits OFF			±4		±4		
Differential Nonlinearity	Over full temp. range		±1	±2		±1	±2	
Settling Time (3)	All Bits ON-to-OFF or OFF-to-ON		300	400		300	400	ns



PARAMETER	CONDITIONS	HI-562A-2/HI-562A-8			HI-562A-4/HI-562A-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
Major Carry Transient Peak Amplitude Settling Time to 90% Complete	From 011...1 to 100...0 or 100...0 to 011...1		0.7			0.7		mA
			35			35		ns
Power Supply Sensitivity (3) Unipolar Offset $V_{ps+}$ @ +5V $V_{ps-}$ @ -15V Bipolar Offset $V_{ps+}$ @ 5V $V_{ps-}$ @ -15V Gain $V_{ps+}$ @ +5V $V_{ps-}$ @ -15V	All Bits OFF		$\pm 0.5$ $\pm 0.5$			$\pm 0.5$ $\pm 0.5$		ppm of FSR/% $V_{ps}$
	All Bits OFF, Bipolar mode		$\pm 1.5$ $\pm 1.5$			$\pm 1.5$ $\pm 1.5$		
	All Bits ON			$\pm 3.5$ $\pm 7.5$			$\pm 3.5$ $\pm 7.5$	

OUTPUT CHARACTERISTICS

Output Current Unipolar Bipolar		-1.6 $\pm 0.8$	-2.0 $\pm 1.0$	-2.4 $\pm 1.2$	-1.6 $\pm 0.8$	-2.0 $\pm 1.0$	-2.4 $\pm 1.2$	mA
Resistance			2K			2K		ohms
Capacitance			20			20		pF
Output Voltage Ranges Unipolar Bipolar	Using external op amp and internal scaling resistors. See Figure 1 and Table 1 for connections		0 to +5 0 to +10 $\pm 2.5$ $\pm 5$ $\pm 10$			0 to +5 0 to +10 $\pm 2.5$ $\pm 5$ $\pm 10$		V
Compliance Limit (3)		-3		+10	-3		+10	V
Compliance Voltage (3)	Over full temp. range		$\pm 1.0$			$\pm 1.0$		V
Output Noise	0.1 to 10Hz (All Bits ON) 0.1 to 5MHz (All Bits ON)		30 100			30 100		$\mu V$ (p-p)

POWER REQUIREMENTS

$V_{ps+}$ (7) $V_{ps-}$	Over full temp. range	4.5 -13.5	5 -15	16.5 -16.5	4.75 -13.5	5 -15	16.5 -16.5	V
$I_{ps+}$ (5) $I_{ps-}$ (5)	All Bits ON or OFF in either TTL or CMOS mode (25°C)		8 16	15 23		8 16	15 23	mA
$I_{ps+}$ (5) $I_{ps-}$ (5)	Same as above except over full temp. range		11 20	20 30		11 20	20 30	mA
Power Dissipation	+25°C $V_{ps+} = +5V$ $V_{ps-} = -15V$		280	420				mW

## High Speed Monolithic Digital to Analog Converter with Reference

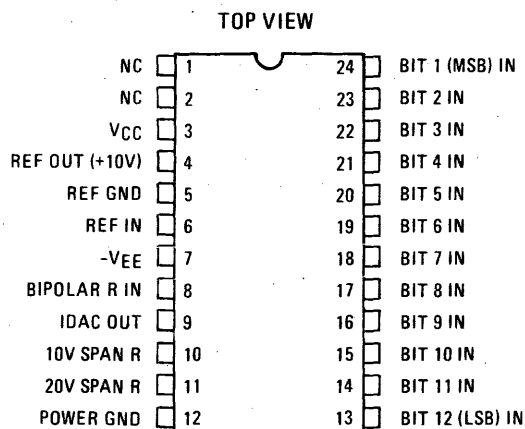
### FEATURES

- DAC AND REFERENCE ON A SINGLE CHIP
- PIN COMPATIBLE WITH AD565A
- VERY HIGH SPEED: SETTLES TO 1/2 LSB IN 250ns, MAX.  
FULL SCALE SWITCHING TIME 30ns, TYP.
- GUARANTEED FOR OPERATION WITH  $\pm 12V$  SUPPLIES
- MONOTONICITY GUARANTEED OVER TEMPERATURE
- 1/2 LSB MAX NONLINEARITY GUARANTEED OVER TEMPERATURE
- LOW GAIN DRIFT (MAX, DAC PLUS REFERENCE) 25ppm/°C
- LOW POWER DISSIPATION 250mW

### APPLICATIONS

- CRT DISPLAYS
- HIGH SPEED A/D CONVERTERS
- SIGNAL RECONSTRUCTION
- WAVEFORM SYNTHESIS

### PINOUT



### DESCRIPTION

The HI-565A is a fast, 12 bit current output, digital to analog converter. The monolithic chip includes a precision voltage reference, thin-film R-2R ladder, reference control amplifier and twelve high-speed bipolar current switches.

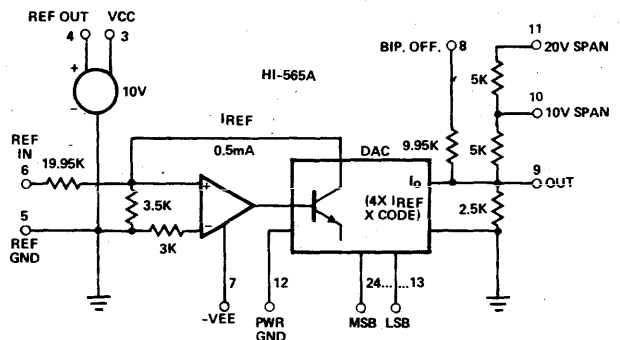
The Harris Semiconductor dielectric isolation process provides latch-free operation while minimizing stray capacitance and leakage currents, to produce an excellent combination of speed and accuracy. Also, ground currents are minimized to produce a low and constant current through the ground terminal, which reduces error due to code-dependent ground currents.

HI-565A dice are laser trimmed for a maximum integral non-linearity error of  $\pm 1/4$  LSB at  $+25^\circ C$ . In addition, the low noise buried zener reference is trimmed both for absolute value and minimum temperature coefficient.

The HI-565A is offered in both commercial and military grades. For high-reliability requirements, additional 100% screening per Mil-Std. 883, Method 5004, Class B is available. See Ordering Information.

Package is a 24 pin side-braced ceramic or plastic DIP. Power requirement is 250mW, typical with  $\pm 15V$  supplies.

### FUNCTIONAL DIAGRAM





**ABSOLUTE MAXIMUM RATINGS\***

V <sub>CC</sub> to Power Ground	0V to +18V	10V Span R to Reference Ground	±12V
V <sub>EE</sub> to Power Ground	0V to -18V	20V Span R to Reference Ground	±24V
Voltage on DAC Output (Pin 9)	-3V to +12V	Ref Out	Indefinite Short to Power Ground Momentary Short to V <sub>CC</sub>
Digital Inputs (Pins 13-24) to Power Ground	-1V to +7.0V	Package Power Dissipation	
Ref In to Reference Ground	±12V	Ceramic (D)	1000mW
Bipolar Offset to Reference Ground	±12V	Plastic (N)	750mW

\*Absolute maximum ratings are limiting values beyond which the serviceability of the circuit may be impaired.

**ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = +25°C, V<sub>CC</sub> = +15V, V<sub>EE</sub> = -15V, Unless Otherwise Specified)

MODEL	HI-565AJ, HI-565AS			HI-565AK, HI-565AT			UNITS
	MIN	TYP	MAX	MIN	TYP	MAX	
<b>DATA INPUTS</b> (Note 1) (Pins 13 to 24)							
<b>TTL or 5V CMOS (T<sub>MIN</sub> to T<sub>MAX</sub>)</b>							
Input Voltage							
Bit ON Logic "1"	+2.0		+5.5	+2.0		+5.5	V
Bit OFF Logic "0"			+0.8			+0.8	V
Logic Current (Each Bit)							μA
Bit ON Logic "1"		.01	+1.0		.01	+1.0	μA
Bit OFF Logic "0"		-2.0	-20		-2.0	-20	μA
<b>RESOLUTION</b>			12			12	Bits
<b>OUTPUT</b>							
Current Unipolar (All Bits On)	-1.6	-2.0	-2.4	-1.6	-2.0	-2.4	mA
Bipolar (All Bits on or Off)	±0.8	±1.0	±1.2	±0.8	±1.0	±1.2	mA
Resistance (Exclusive of Span Resistors)	1.8k	2.5k	3.2k	1.8k	2.5k	3.2k	Ω
Offset Unipolar		0.01	0.05		0.01	0.05	% of F.S.
Bipolar (Figure 2, R <sub>3</sub> = 50Ω Fixed)		0.05	0.15		0.05	0.1	% of F.S.
Capacitance		20			20		pF
Compliance Voltage, T <sub>MIN</sub> to T <sub>MAX</sub>	-1.5		+10	-1.5		+10	V
<b>ACCURACY</b> (Error Relative to Full Scale)							
+25°C		±1/4 (0.006)	±1/2 (0.012)		±1/8 (0.003)	±1/4 (0.006)	LSB % of F.S.
T <sub>MIN</sub> to T <sub>MAX</sub>		±1/2 (0.012)	±3/4 (0.018)		±1/4 (0.006)	±1/2 (0.012)	LSB % OF F.S.
<b>DIFFERENTIAL NONLINEARITY</b>							
+25°C		±1/2	±3/4		±1/4	±1/2	LSB
T <sub>MIN</sub> to T <sub>MAX</sub>	MONOTONICITY GUARANTEED						
<b>TEMPERATURE COEFFICIENTS</b>							
With Internal Reference							
Unipolar Zero		1	2		1	2	ppm/°C
Bipolar Zero		5	10		5	10	ppm/°C
Gain (Full Scale)		15	40		10	25	ppm/°C
Differential Nonlinearity		2			2		ppm/°C
<b>SETTLING TIME TO 1/2 LSB</b>							
With High Z External Load (Note 2)		350	500		350	500	ns
With 75Ω External Load		150	250		150	250	ns





MODEL	HI-565AJ, HI-565AS			HI-565AK, HI-565AT			UNITS
	MIN	TYP	MAX	MIN	TYP	MAX	
<b>FULL SCALE TRANSITION (From 50% of Logic Input to 90% of Analog Output)</b>							
Rise Time		15	30		15	30	ns
Fall Time		30	50		30	50	ns
<b>TEMPERATURE RANGE</b>							
Operating (HI-565AJ/K)	0		+75	0		+75	°C
(HI-565AS/T)	-55		+125	-55		+125	°C
Storage							
D Package (All)	-65		+150	-65		+150	°C
N Package (J, K)	-25		+150	-25		+150	°C
<b>POWER REQUIREMENTS</b>							
V <sub>CC</sub> , +11.4 to +16.5V DC		7.0	10.5		7.0	10.5	mA
V <sub>EE</sub> , -11.4 to -16.5V DC		-9.5	-14.5		-9.5	-14.5	mA
<b>POWER SUPPLY GAIN SENSITIVITY (Note 3)</b>							
V <sub>CC</sub> = +11.4 to +16.5 VDC		3	10		3	10	ppm of F.S./%
V <sub>EE</sub> = -11.4 to -16.5 VDC		15	25		15	25	ppm of F.S./%
<b>PROGRAMMABLE OUTPUT RANGES (See Table 1)</b>							
		0 to +5			0 to +5		V
		-2.5 to +2.5			-2.5 to +2.5		V
		0 to +10			0 to +10		V
		-5 to +5			-5 to +5		V
		-10 to +10			-10 to +10		V
<b>EXTERNAL ADJUSTMENTS</b>							
Gain Error with Fixed 50Ω Resistor for R2 (Figure 1)		±0.1	±0.25		±0.1	±0.25	% of F.S.
Bipolar Zero Error with Fixed 50Ω Resistor for R3 (Figure 2)		±0.05	±0.15		±0.05	±0.1	% of F.S.
Gain Adjustment Range (Figure 1)	±0.25			±0.25			% of F.S.
Bipolar Zero Adjustment Range	±0.15			±0.15			% of F.S.
<b>REFERENCE INPUT</b>							
Input Impedance	15K	20K	25K	15K	20K	25K	
<b>REFERENCE OUTPUT</b>							
Voltage	9.90	10.00	10.10	9.90	10.00	10.10	V
Current (Available for External Loads)	1.5	2.5		1.5	2.5		mA
<b>POWER DISSIPATION</b>							
		250	375		250	375	mW

**NOTES:**

1. Guaranteed but not tested over the operating temperature range.
2. See settling time discussion and Figure 3.
3. The Power Supply Gain Sensitivity is tested in reference to a V<sub>CC</sub>, V<sub>EE</sub> of ±15V.

## 10 Bit High Speed Monolithic Digital-to-Analog Converter

### FEATURES

- MONOLITHIC CONSTRUCTION
- EXTREMELY FAST SETTLING. . . . . 85ns TO ½LSB TYP.
- LOW GAIN DRIFT. . . . . ±5ppm/°C TYP.
- EXCELLENT LINEARITY OVER TEMPERATURE . . . . . ± ½LSB MAX.
- DESIGNED FOR MINIMUM GLITCHES
- MONOTONIC OVER TEMPERATURE

### APPLICATIONS

- CRT DISPLAY GENERATION
- HIGH SPEED A/D CONVERTERS
- VIDEO SIGNAL RECONSTRUCTION
- WAVEFORM SYNTHESIZERS
- HIGH SPEED DATA ACQUISITION
- HIGH RELIABILITY APPLICATIONS
- PRECISION INSTRUMENTS

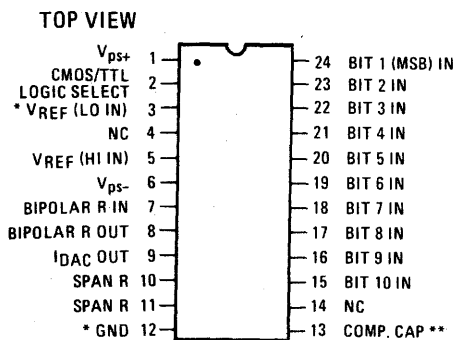
### DESCRIPTION

The HI-5610 is an ultra-high speed 10 bit monolithic current output digital-to-analog converter. The fast output current settling of 85ns to ½LSB of its final value is achieved using dielectric isolation processing to reduce internal parasitics for fast rise and fall times during switching. Output glitches are minimized in the HI-5610 by incorporating equally weighted current sources switched into an R-2R ladder network for symmetrical turn-on and turn-off switching times. This creates within the chip a very uniform and constant thermal distribution for excellent linearity and also eliminates thermal transients during switching. High stability thin film resistor processing, together with laser trimming provide the HI-5610 with true 10 bit linearity to within ± ½LSB maximum over operating temperature range. The HI-5610's low offset and gain drift over the operating temperature range assures that its absolute accuracy when referred to a fixed 10V reference will not deviate more than ± 1LSB for both unipolar and bipolar operation.

The HI-5610 is recommended as a replacement for high cost hybrid and modular units for increased reliability and accuracy in applications such as CRT Displays, precision instruments and data acquisition system requiring through-put rates as high as 12MHz for full range transitions. Its small size makes it an ideal choice as the essential part of high speed A/D converter designs or as a building block in high speed or high resolution industrial process control systems. The HI-5610 is also ideally suited for aircraft and space instrumentation where operation over a wide temperature range is required.

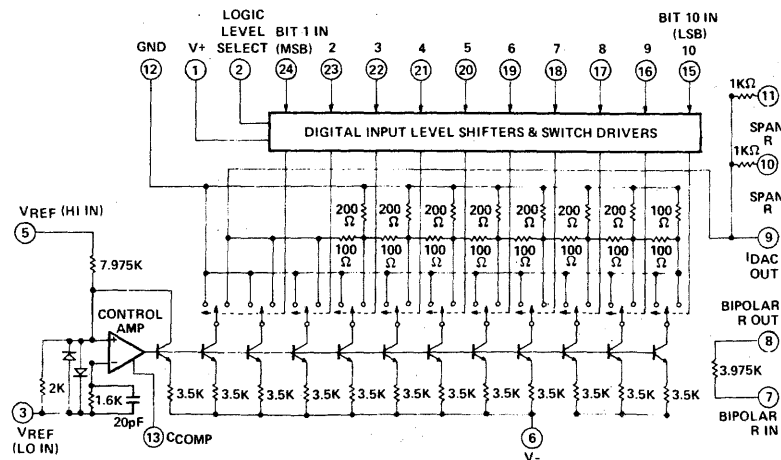
The HI-5610-5 is specified for operation over 0°C to +75°C, the HI-5610-2 and HI-5610-8 over -55°C to +125°C. Processing to MIL-STD-883A class B screening is available by selecting the HI-5610-8. All are available in a hermetically sealed 24 lead dual-in-line package.

### PINOUT



\* Pin 3 connected to bottom case for high frequency shielding.  
 \*\* For high speed operation, connect 0.01 μF between Pin 13 and GND. Otherwise, leave Pin 13 open.

### FUNCTIONAL DIAGRAM





**ABSOLUTE MAXIMUM RATINGS** (Referred to Ground)<sup>1</sup>

Power Supply Inputs	$V_{ps+}$	+20V	Power Dissipation $P_d$ , Package	1000mW
	$V_{ps-}$	-20V		
Reference Inputs	$V_{REF}$ (Hi)	$\pm V_{ps}$	Operating Temperature Range	
	$V_{REF}$ (Lo)	0V	HI-5610-2	-55°C to +125°C
			HI-5610-5	0°C to +75°C
Digital Inputs	Bits 1 - 12	-1V, +12V	HI-5610-8	-55°C to +125°C
	CMOS/TTL Logic Select	-1V, +12V		
Outputs	Pins 7, 8, 10, 11	$\pm V_{ps}$	Storage Temperature Range	-65°C to +150°C
	Pin 9	+ $V_{ps}$ , -5V		

**ELECTRICAL CHARACTERISTICS** (@ +25°C,  $V_{ps+}$  = +5V,  $V_{ps-}$  = -15V,  $V_{REF}$  = +10V, pin 2 ground unless otherwise noted)

PARAMETER	TEMP	HI-5610-2 HI-5610-8			HI-5610-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Digital Inputs (2)								
TTL Logic Input Voltage (3)								
Logic "1"	Full	2.0			2.0			V
Logic "0"	Full			0.8			0.8	V
Input Current								
Logic "1"	Full		20	500		20	500	nA
Logic "0"	Full		-50	-100		-50	-100	μA
CMOS Logic Input Voltage (4)								
Logic "1"	Full	0.7 $V_{ps+}$			0.7 $V_{ps+}$			V
Logic "0"	Full			0.3 $V_{ps+}$			0.3 $V_{ps+}$	V
Input Current								
Logic "1"	Full		20	500		20	500	nA
Logic "0"	Full		-50	-100		-50	-100	μA
Reference Input								
Input Resistance			8K			8K		Ω
Input Voltage ( $I_{OUT}$ = 5mA + 20%)			+10			+10		V
<b>TRANSFER CHARACTERISTICS</b>								
Resolution	Full			10			10	Bits
Nonlinearity (5)	25°C			± ½			± ½	LSB
	25°C			± ½			± ½	LSB
Relative Accuracy (6)								(9)
Gain Error (Input Code 11...1)				± 0.05			± 0.05	% FSR
Unipolar Offset Error (Input Code 00...0)				± 0.05			± 0.05	% FSR
Bipolar Offset Error (Input Code 00...0) (Adjustable to zero, see Figure 4, 5)				± 0.05			± 0.05	% FSR
Adjustment Range								
Gain				± 0.25			± 0.25	% FSR
Bipolar Offset				± 0.25			± 0.25	% FSR
Temperature Stability								
Gain Drift	Full			± 5			± 5	ppm/°C
Unipolar Offset Drift	Full			± 3			± 3	ppm/°C
Bipolar Offset Drift	Full			± 3			± 3	ppm/°C
Differential Nonlinearity	Full			± 2			± 2	ppm/°C
<b>MONOTONICITY - GUARANTEED OVER FULL OPERATING TEMPERATURE RANGE</b>								
Settling Time to ½LSB (5)								
From all 0's to all 1's						85		ns
From all 1's to all 0's						85		ns
Major Carry Switching to 90% Complete			40			40		ns



PARAMETER	TEMP	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Power Supply Sensitivity (5) $V_{ps+} = +5V, V_{ps-} = -13.5V$ to $-16.5V$ Gain (Input Code 11....1) Unipolar Offset (Input Code 00....0) Bipolar Offset (Input Code 00....0)				$\pm 3.5$		$\pm 0.5$	$\pm 3.5$	ppm of FSR/% $V_{ps}$
$V_{ps-} = -15V, V_{ps+} = 4.5V$ to $5.5V$ Gain (Input Code 11....1) Unipolar Offset (Input Code 00....0) Bipolar Offset (Input Code 00....0)				$\pm 7.5$	$\pm 0.5$	$\pm 7.5$		
OUTPUT CHARACTERISTICS								
Output Current Unipolar Bipolar		-4.0 $\pm 2.0$	-5.0 $\pm 2.5$	-6.0 $\pm 3.0$	-4.0 $\pm 2.0$	-5.0 $\pm 2.5$	-6.0 $\pm 3.0$	mA mA
Output Resistance			200			200		$\Omega$
Output Capacitance			20			20		pF
Output Voltage Range (7) Unipolar Bipolar			+5 +2.5 $\pm 2.5$ $\pm 1.25$			+5 +2.5 $\pm 2.5$ $\pm 1.25$		V V V V
Output Compliance Limit (5)		-3		+10	-3		+10	V
Output Compliance Voltage (5)	Full		$\pm 1.5$			$\pm 1.5$		V
Output Noise Voltage (8) 0.1Hz to 100Hz 0.1Hz to 1MHz			10 100			10 100		$\mu V_{p-p}$ $\mu V_{p-p}$
POWER REQUIREMENTS								
$V_{ps+}$ (4)	Full	4.5	5	16.5	4.75	5	16.5	V
$V_{ps-}$	Full	-13.5	-15	-16.5	-13.5	-15	-16.5	V
$I_{ps+}$ (All 1's or all 0's in (10) either TTL or CMOS Mode)	25°C Full		9 20			9 20		mA mA
$I_{ps-}$ (Same as above) (10)	25°C Full		25 30			25 30		mA mA

NOTES:

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. The HI-5610 accepts digital input codes in binary format and may be user connected for any one of three binary codes. Straight binary, offset binary, or two's complement binary. (See operating instructions).
3. For TTL and DTL compatibility connect +5V to pin 1 and ground pin 2. The  $V_{ps+}$  tolerance is  $\pm 10\%$  for HI-5610-2,-8. And  $\pm 5\%$  for HI-5610-5.
4. For CMOS compatibility based on  $V_{ps+} \geq +8V$ , (switching thresholds equal  $V_{ps+}/2$ ), connect pins 1 and 2. For CMOS levels below +8V, connect pin 2 to ground only (this provides a threshold of approximately +1.4V).
5. See definitions.
6. Using an external op amp with internal span resistors and 24.9 $\Omega$   $\pm 1\%$  external trim resistors in place of potentiometers R1 and R2. These errors are adjustable to zero using R1 and R2. (See operating instructions.)
7. Using an external op amp and internal span resistors. (See operating instructions for connections.)
8. Specified for digital input in all '1's or all '0's.
9. FSR is "Full Scale Range" and is 5V for  $\pm 2.5V$  range, 2.5V for  $\pm 1.25V$  range, etc., or 5mA ( $\pm 20\%$ ) for current output.
10. After 30 seconds warm-up.



# HI-5618A/5618B

## 8 Bit High Speed Digital-to-Analog Converters

INTERFACE

Harris Semiconductor

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• VERY FAST SETTLING CURRENT OUTPUT 65ns</li> <li>• MINIMUM NONLINEARITY ERROR               <ul style="list-style-type: none"> <li>HI-5618A <math>\pm 1/4</math> LSB MAX</li> <li>HI-5618B <math>\pm 1/2</math> LSB MAX</li> </ul> </li> <li>• LOW POWER OPERATION 340mW TYP</li> <li>• ON-CHIP RESISTORS FOR GAIN AND BIPOLAR OFFSET</li> <li>• GUARANTEED MONOTONIC OVER TEMPERATURE</li> <li>• CMOS, TTL, OR DTL COMPATIBLE</li> </ul>	<p>The HI-5618A/B are very high speed 8 bit current output D/A converters. These monolithic devices are fabricated with dielectrically isolated bipolar processing, which reduces internal parasitic capacitance to allow fast rise and fall times. This achieves a typical full scale settling time of 65ns to <math>\pm 1/2</math> LSB. Output glitches are minimized by incorporation of equally weighted current sources, switched to either an R-2R ladder network or ground for symmetrical turn ON and turn OFF times. High stability thin film resistors provide excellent accuracy without trimming. For example, the HI-5618A has <math>\pm 1/4</math> LSB maximum nonlinearity error at +25°C, with <math>\pm 3/8</math> LSB guaranteed over the full operating temperature range.</p> <p>The HI-5618A/B are recommended for any application requiring high speed and accurate conversions. They can be used in CRT displays and systems requiring throughput rates as high as 20MHz for full scale transitions. Other applications include high speed process control, defense systems, avionics, and space instrumentation.</p> <p>The HI-5618A-5 and HI-5618B-5 are specified for operation from 0°C to +75°C. The "-2" versions are specified from -55°C to +125°C. "Dash 8" (-8) designates parts which have been screened per MIL-STD-883, Method 5004/Class B.</p> <p>Power requirements are +5V and -15V. Package is an 18 pin DIP, in plastic or ceramic.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• HIGH SPEED PROCESS CONTROL</li> <li>• CRT DISPLAY GENERATION</li> <li>• HIGH SPEED A/D CONVERSION</li> <li>• WAVEFORM SYNTHESIS</li> <li>• HIGH RELIABILITY APPLICATIONS</li> <li>• VIDEO SIGNAL RECONSTRUCTION</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;">TOP VIEW</p>	

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**ABSOLUTE MAXIMUM RATINGS** (Referred to Ground) (1)

Power Supply Inputs	$V_{ps+}$	+20V	Power Dissipation Pd, Package	700mW
	$V_{ps-}$	-20V	Operating Temperature Range	
Reference Inputs	$V_{REF}$ (Hi)	$\pm V_{ps}$	HI-5618A/B-2	-55°C to +125°C
	$V_{REF}$ (Lo)	0V	HI-5618A/B-5	0°C to +75°C
Digital Inputs	Bits 1 - 8	-1V, +12V	HI-5618A/B-8	-55°C to +125°C
CMOS/TTL Logic Select		-1V, +12V	Storage Temperature Range	-65°C to 150°C
Outputs	Pins 5, 7, 8	$\pm V_{ps}$		
	Pin 6	$+V_{ps}, -2.5V$		

**ELECTRICAL CHARACTERISTICS** ( $V_{ps+} = +5V$ ;  $V_{ps-} = -15V$ ;  $V_{REF} = +10V$ ; Pin 2 to GND, unless otherwise noted)

PARAMETER	TEMP	HI-5618A/B-2 HI-5618A/B-8			HI-5618A/B-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

**INPUT CHARACTERISTICS**

Digital Inputs (2)									
TTL logic Input Voltage (3)	Logic "1"	Full	2.0			2.0			V
	Logic "0"	Full			0.8			0.8	V
TTL Logic Input Current	Logic "1"	Full		20	500		20	500	nA
	Logic "0"	Full		-50	-100		-50	-100	$\mu$ A
CMOS Logic Input Voltage (4)	Logic "1"	Full	0.7 $V_{ps+}$			0.7 $V_{ps+}$			V
	Logic "0"	Full			0.3 $V_{ps+}$			0.3 $V_{ps+}$	V
CMOS Logic Input Current	Logic "1"	Full		20	500		20	500	nA
	Logic "0"	Full		-50	-100		-50	-100	$\mu$ A
Reference Input									
Input Resistance		+25°C		8k			8k		$\Omega$
Input Voltage ( $I_{OUT} = 5mA \pm 20\%$ )		+25°C		+10			+10		V

**TRANSFER CHARACTERISTICS**

Resolution		Full		8			8		Bits
Nonlinearity, Integral and Differential	HI-5618A	25°C			$\pm 1/4$			$\pm 1/4$	LSB
		Full			$\pm 3/8$			$\pm 3/8$	LSB
	HI-5618B	25°C			$\pm 1/2$			$\pm 1/2$	LSB
		Full			$\pm 5/8$			$\pm 5/8$	LSB
Initial Accuracy (6) (Relative to External +10V Reference)									
Gain		25°C			$\pm 2$			$\pm 2$	LSB
Unipolar Zero		25°C			$\pm 1/8$			$\pm 1/8$	LSB
Bipolar Offset (Neg. Full Scale)		25°C			$\pm 2$			$\pm 2$	LSB
Temperature Stability									
Gain Drift		Full			$\pm 1/4$			$\pm 1/4$	LSB
Unipolar Zero Drift		Full			$\pm 1/16$			$\pm 1/16$	LSB
Bipolar Zero Drift		Full			$\pm 1/4$			$\pm 1/4$	LSB
Settling Time (5) to 1/2 LSB									
High Impedance (11)	(from all 0's to all 1's) or (from all 1's to all 0's)	+25°C		65	75		65	75	ns



PARAMETER	TEMP	HI-5618A/B-2 HI-5618A/B-8			HI-5618A/B-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

TRANSFER CHARACTERISTICS (Continued)

Glitch (5) - Major Carry Transition									
Duration	+25°C		20				20		ns
Amplitude (See Fig. 4)	+25°C		350				350		mV
Area	+25°C		3500				3500		mV-ns
Power Supply Sensitivity (5)									
$V_{ps+} = +5V, V_{ps-} = -13V$ to $-16.5V$									
Gain (Input Code 11...1)	+25°C			±5				±5	ppm of FSR/% $V_{ps}$ (9)
Unipolar Zero (Input Code 00...0)	+25°C		±0.5			±0.5			
Bipolar Offset (Input Code 00...0)	+25°C		±1.5			±1.5			
$V_{ps-} = -15V, V_{ps+} = 4.5V$ to $5.5V$									
Gain (Input Code 11...1)	+25°C			±5				±5	
Unipolar Zero (Input Code 00...0)	+25°C		±0.5			±0.5			
Bipolar Offset (Input Code 00...0)	+25°C		±1.5			±1.5			

OUTPUT CHARACTERISTICS

Output Current	Unipolar	+25°C	-4	-5	-6	-4	-5	-6	mA
	Bipolar	+25°C	±2.0	±2.5	±3.0	±2.0	±2.5	±3.0	mA
Output Resistance		+25°C		500			500		Ω
Output Capacitance		+25°C		20			20		pF
Output Voltage Range (7)	Unipolar	+25°C		+10			+10		V
		+25°C		+5			+5		V
	Bipolar	+25°C		±10			±10		V
		+25°C		±5			±5		V
		+25°C		±2.5			±2.5		V
Output Compliance Voltage (5)		+25°C		±1.5			±1.5		V
Output Noise Voltage (8)	0.1Hz to 100Hz	+25°C		30			30		μV <sub>p-p</sub>
	0.1Hz to 1Mhz	+25°C		100			100		μV <sub>p-p</sub>

POWER REQUIREMENTS (4)

$V_{ps+}$	Full	4.5	5	16.5	4.75	5	16.5	V
$V_{ps-}$	Full	-13.5	-15	-16.5	-14.25	-15	-15.75	V
$I_{ps+}$ (10) (All 1's or all 0's in either TTL or CMOS mode) (3, 4)	+25°C		9			9		mA
	Full			12			12	mA
$I_{ps-}$ (10) (All 1's or all 0's in either TTL or CMOS mode) (3, 4)	+25°C		19			19		mA
	Full			26			26	mA

- Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
- The HI-5618 accepts digital input codes in binary format and may be user connected for any one of three binary codes. Straight binary, offset binary, or two's complement binary. (See operating instructions)
- For TTL and DTL compatibility connect +5V to pin 1 and ground pin 2. The  $V_{ps+}$  tolerance is ±10% for HI-5618A/B-2, -8; and ±5% for HI-5618A/B-5.
- For CMOS compatibility based on  $V_{ps+} \geq +8V$ , (switching thresholds equal  $V_{ps+}/2$ ), connect pins 1 and 2. For CMOS levels below +8V, connect pin 2 to ground only (this provides a threshold of approximately +1.4V).
- See definitions.
- These errors may be adjusted to zero using external potentiometers  $R_1, R_2, R_3$ .  $R_1$  and  $R_2$  each provide more than ±3 LSB's adjustment. (See Operating Instructions). The specifications listed under initial accuracy are based on use of an external op amp, internal span and offset resistors, and 100Ω ±1% resistors, in place of  $R_1$  and  $R_2$ .
- Using an external op amp with the internal span and offset resistors. See Operating Instructions.
- Specified for all "1's" or all "0's" digital input.
- FSR is "Full Scale Range", i.e., 20V for ±10V range; 10V for ±5V range, etc. Nominal full scale output current is 5mA.
- After 30 seconds warm-up.
- See Test Circuit, Figure 3.
- See Test Circuit, Figure 4.

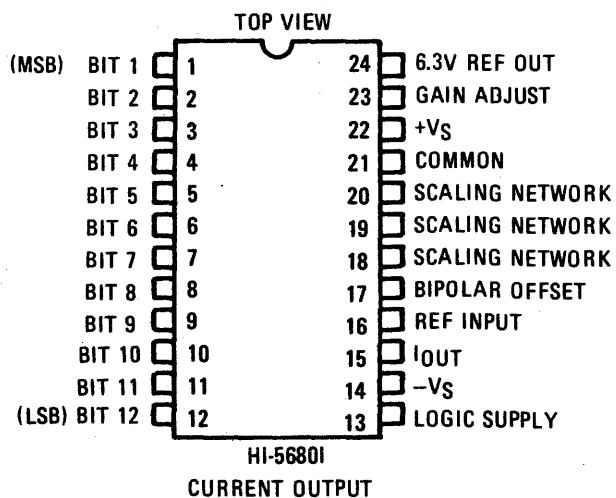
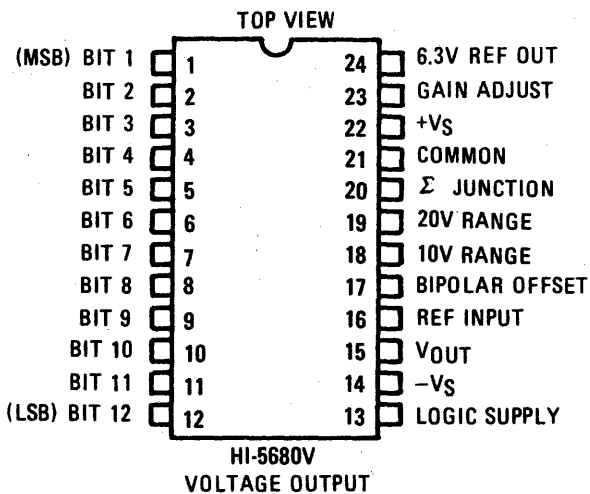


# HI-5680

## 12 Bit Low Cost Monolithic Digital-to-Analog Converter

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• DAC 80 ALTERNATE SOURCE</li> <li>• MONOLITHIC CONSTRUCTION (SINGLE CHIP)</li> <li>• FAST SETTLING</li> <li>• GUARANTEED MONOTONIC <span style="float: right;">0°C to 75°C</span></li> <li>• WAFER LASER TRIMMED</li> <li>• APPLICATIONS RESISTORS ON-CHIP</li> <li>• ON-BOARD REFERENCE</li> <li>• DIELECTRIC ISOLATION (DI) PROCESSING</li> <li>• ±12V POWER SUPPLY OPERATION</li> </ul>	<p>The HI-5680 is a monolithic, direct replacement for the popular DAC80-CBI, DAC80Z-CBI, and DAC85C-CBI, incorporating the best features of each. Single chip construction, along with several design innovations, make the HI-5680 the optimum choice for low cost, high reliability applications.</p> <p>Harris' unique Dielectric Isolation(DI) processing reduces internal parasitics, resulting in fast switching times and minimum glitch. On-board span resistors are provided for good tracking over temperature, and are laser trimmed to high accuracy. These may be used with the on-board op-amp (voltage output models; HI-5680V), or with a user supplied external amplifier (HI-5680I).</p> <p>Internally, the HI-5680 eliminates code dependent ground currents by routing current from the positive supply to the internal ground node, as determined by an auxiliary R-2R ladder. This results in a cancellation of code dependent ground currents allowing virtually zero variation in current through the package common, pin 21.</p> <p>The HI-5680 is available in both current and voltage output models which are guaranteed over the 0°C to +75°C temperature range. All models include a buried zener reference featuring low temperature coefficient. In addition, the voltage output models include an on-board output amplifier. Both versions operate with a +5V logic supply and a ±V<sub>S</sub> in the range of ±(11.4V to 16.5V).</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• HIGH SPEED A/D CONVERTERS</li> <li>• PRECISION INSTRUMENTATION</li> <li>• CRT DISPLAY GENERATION</li> </ul>	

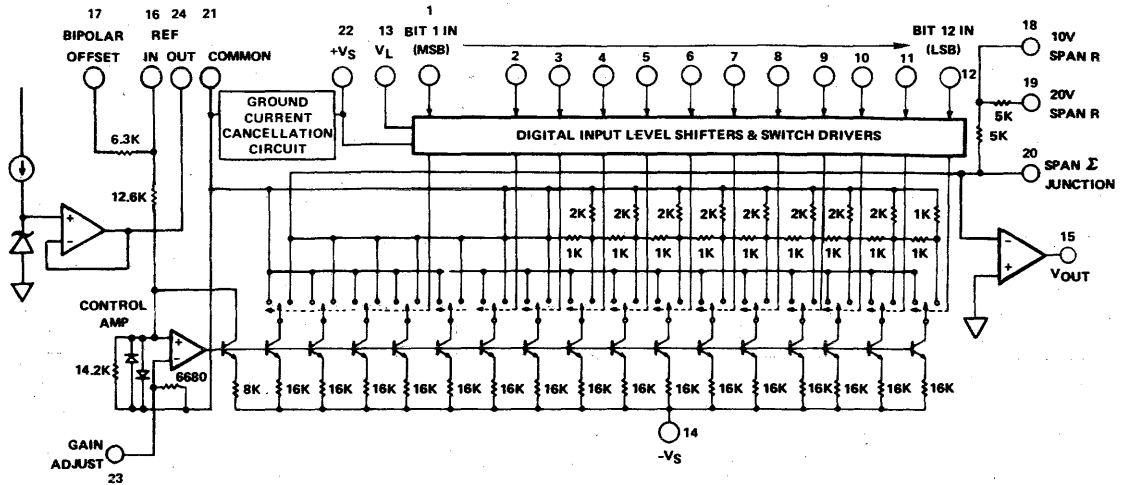
### PINOUTS



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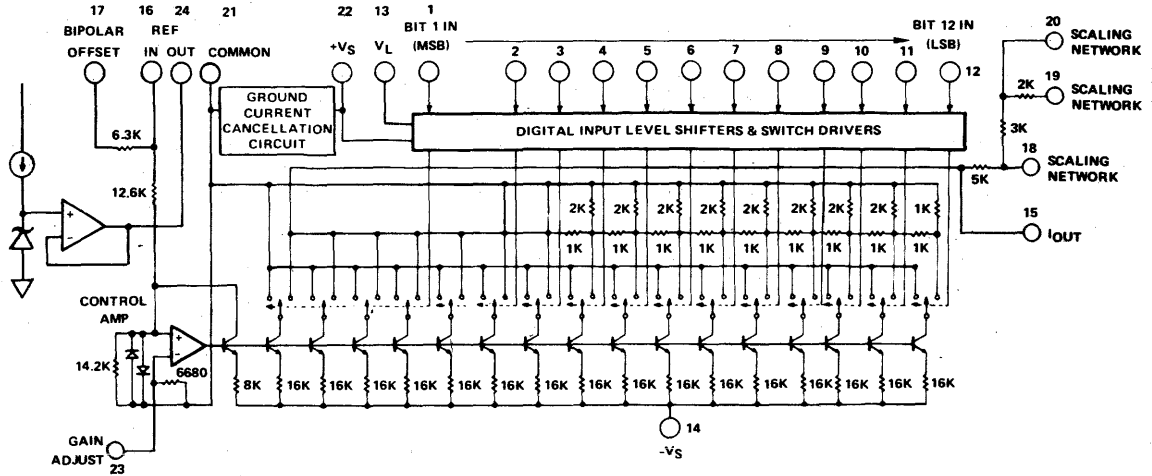


**FUNCTIONAL DIAGRAM VOLTAGE OUTPUT**



**HI-5680 V**

**FUNCTIONAL DIAGRAM CURRENT OUTPUT**



**HI-5680 I**

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (1)

Power Supply Inputs	+V <sub>S</sub>	+20V	Power Dissipation	Pd, Package	1000mW
	-V <sub>S</sub>	-20V	Operating Temperature Range	HI-5680I/V-5	0°C to +75°C
	+V <sub>LOGIC</sub>	+20V	Storage Temperature Range		-65°C to +150°C
Reference	Input (pin 16)	+V <sub>S</sub>			
	Output drain	2.5mA			
Digital Inputs	Bits 1 to 12	-1V to +12V			

## ELECTRICAL SPECIFICATIONS

(T<sub>A</sub> = +25°C, V<sub>S</sub> = ±15V, V<sub>LOGIC</sub> = +5V, PIN 16 CONNECTED TO PIN 24 UNLESS OTHERWISE SPECIFIED.)

PARAMETER	CONDITIONS	HI-5680X			UNITS
		MIN	TYP	MAX	
<b>DIGITAL INPUT (3)</b>					
Resolution	TTL Compatible			12	Bits
Logic Levels	at +1 μA	+2		+5.5	Volts
Logic "1"	at -100 μA	0		+0.8	Volts
Logic "0"					
<b>ACCURACY (3)</b>					
Linearity Error	0°C to +75°C		±¼	±½	LSB
Differential Lin. Error	0°C to +75°C		±½	±¾	LSB
Gain Error (2)			±0.1	±0.3	%FSR
Offset Error (2)			±0.05	±0.15	%FSR
Monotonicity	0°C to +75°C		Guaranteed		
<b>DRIFT (3)</b>					
Total Bipolar Drift (Includes gain, offset and linearity drifts.)	0°C to +75°C			±20	PPM/°C
Total Error	0°C to +75°C				
Unipolar			±0.08	±0.15	%FSR
Bipolar			±0.06	±.1	%FSR
Gain	Including internal reference		±15	±30	PPM/°C
	Exclusive of internal reference		±5	±7	PPM/°C
Unipolar Offset			±1	±3	PPM/°C
Bipolar Offset			±5	±10	PPM/°C
<b>CONVERSION SPEED (3)</b>					
Voltage Models					
Settling time (3)	to ±0.01% of FSR for FSR Change				
With 10KΩ Feedback			3		μs
With 5KΩ Feedback			1.5		μs
For 1 LSB change			1.5		μs
Slew Rate		10	15		V/μs
Current Models					
Settling time (3)	to ±0.01% of FSR for FSR Change				
10 to 100Ω load			300		ns
1KΩ load			1000		ns
<b>ANALOG OUTPUT</b>					
Voltage Models					
Output current		±5			mA
Output Resistance			.05		Ω
Short Circuit Duration	to common		continuous		



PARAMETER	CONDITIONS	HI-5680X			UNITS
		MIN	TYP	MAX	
<b>ANALOG OUTPUT</b>					
Current Models					
Output Current					
Unipolar		-1.6	-2	-2.4	mA
Bipolar		±0.8	±1	±1.2	mA
Output Impedance					
Unipolar			2.0		KΩ
Bipolar			2.0		KΩ
Compliance (3)				±2.5	V
<b>INTERNAL REFERENCE</b>					
Output Voltage		+6.174	+6.3	+6.426	V
Output Impedance			1.5		Ω
External Current				+2.5	mA
Tempco of Drift			20		PPM/°C
<b>POWER SUPPLY SENSITIVITY (3)</b>					
+15V supply				.002	%FSR
-15V supply				.002	$\frac{\Delta V_s}{V_s}$
+5V supply				.002	
<b>POWER SUPPLY REQUIREMENTS (5)</b>					
Range					
+15V		+11.4	+15	+16.5	V
-15V		-11.4	-15	-16.5	V
+5V		+ 4.5	+ 5	+16.5	V
Current					
+15V			8	11	mA
-15V			-12	-20	mA
+5V			4.5	8	mA

NOTES:

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. Adjustable to zero using external potentiometers.
3. See definitions.
4. FSR is "full scale range" and is 20V for ±10V range, 10V for ±5V range, etc., or 2mA (±20%) for current output.
5. The HI-5680 will operate with supply voltages as low as ±11.4V. It is recommended that output voltage range -10V to +10V not be used if the supply voltages are less than ±12.5V.

## DEFINITIONS OF SPECIFICATIONS



### DIGITAL INPUTS

The HI-5680 accepts digital input codes in complementary binary, complementary offset binary, and complementary two's complement binary.

DIGITAL INPUT	ANALOG OUTPUT		
	Complementary Binary	Complementary Offset Binary	Complementary Two's Complement*
MSB LSB 000...000	+ Full Scale	+ Full Scale	- LSB
100...000	Mid Scale-1LSB	-1 LSB	+ Full Scale
111...111	Zero	- Full Scale	Zero
011...111	+½ Full Scale	Zero	- Full Scale

\*Invert MSB with external inverter to obtain CTC Coding

### SETTLING TIME

That interval between application of a digital step input, and final entry of the analog output within a specified window about the settled value. Harris Semiconductor usually specifies a unipolar 10V full scale step, to be measured from 50% of the input digital transition, and a window of  $\pm 1/2$  LSB about the final value. The device output is then rated according to the worst (longest settling) case: low to high, or high to low.

### DRIFT

**GAIN DRIFT** – The change in full scale analog output over the specified temperature range expressed in parts per million of full scale per °C (ppm of FSR/°C). Gain error is measured with respect to +25°C at high ( $T_H$ ) and low ( $T_L$ ) temperatures. Gain drift is calculated for both high ( $T_H - 25^\circ\text{C}$ ) and low ranges (+25°C- $T_L$ ) by dividing the gain error by the respective change in temperature. The specification is the larger of the two representing worst case drift.

**OFFSET DRIFT** – The change in analog output with all bits OFF over the specified temperature range expressed in parts per million of full scale range per °C (ppm of FSR/°C). Offset error is measured with respect to +25°C at high ( $T_H$ ) and low ( $T_L$ ) temperatures. Offset Drift is calculated for both high ( $T_H - 25^\circ\text{C}$ ) and low (+25°C- $T_L$ ) ranges by dividing the offset error by the respective change in temperature. The specification given is the larger of the two, representing worst-case drift.

### ACCURACY

**INTEGRAL NONLINEARITY** – The maximum deviation of the actual transfer characteristic from an ideal straight line. The ideal line is positioned according to "end-point linearity" for D/A converter products from Harris Semiconductor, i.e. the line is drawn between the end-points of the actual transfer characteristic (codes 00...0 and 11...1).

**DIFFERENTIAL NONLINEARITY** – The difference between one LSB and the output voltage change corresponding to any two consecutive codes. A Differential Nonlinearity of  $\pm 1$  LSB or less guarantees monotonicity.

**MONOTONICITY** – The property of a D/A converter's transfer function which guarantees that the output derivative will not change sign in response to a sequence of increasing (or decreasing) input codes. That is, the only output response to a code change is to remain constant, increase for increasing code, or decrease for decreasing code.

### POWER SUPPLY SENSITIVITY

Power Supply Sensitivity is a measure of the change in gain and offset of the D/A converter resulting from a change in -15V, or +15V supplies. It is specified under DC conditions and expressed as parts per million of full scale range per percent of change in power supply (ppm of FSR/%).

### COMPLIANCE

Compliance voltage is the maximum output voltage range that can be tolerated and still maintain its specified accuracy. Compliance limit implies functional operation only and makes no claims to accuracy.

### GLITCH

A glitch on the output of a D/A converter is a transient spike resulting from unequal internal ON-OFF switching times. Worst case glitches usually occur at half-scale or the major carry code transition from 011...1 to 100...0 or vice versa. For example, if turn ON is greater than turn OFF for 011...1 to 100...0, an intermediate state of 000...0 exists, such that, the output momentarily glitches toward zero output. Matched switching times and fast switching will reduce glitches considerably. (Measured as one half the product of duration and amplitude.)

**DECOUPLING AND GROUNDING**

For best accuracy and high frequency performance, the grounding and decoupling scheme shown in Figure 1 should be used. Decoupling capacitors should be connected close to the HI-5680 (preferably to the device pins) and should be tantalum or electrolytic bypassed with ceramic types for best high frequency noise rejection.

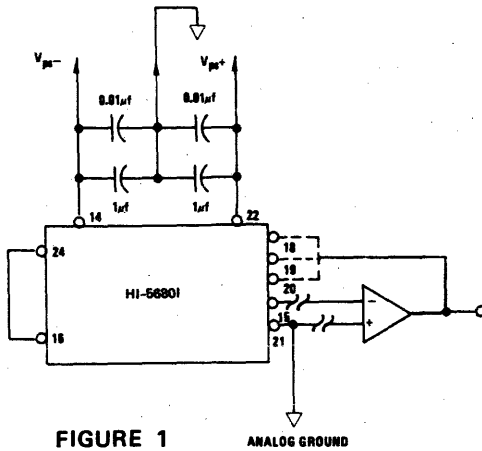


FIGURE 1

**REFERENCE SUPPLY**

An internal 6.3Volt reference is provided on board all HI-5680 models. This voltage (pin 24) is accurate to  $\pm 2\%$  and must be connected to the reference input (pin 16) for specified operation. This reference may be used externally, provided current drain is limited to 2.5mA. An external buffer amplifier is recommended if this reference is to be used to drive other system components. Otherwise, variations in the load driven by the reference will result in gain variations of the HI-5680. All gain adjustments should be made under constant load conditions.

**VOLTAGE OUTPUT HI-5680V**

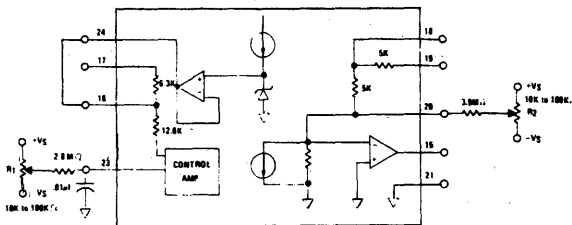


FIGURE 2

**RANGE CONNECTIONS**

	Range	Connect		
		PIN 15	PIN 17	PIN 19
Unipolar	0 to +5V	18	N.C.	20
	0 to +10V	18	N.C.	N.C.
Bipolar	$\pm 2.5V$	18	20	20
	$\pm 5V$	18	20	N.C.
	$\pm 10V$	19	20	15

**CURRENT OUTPUT HI-5680I**

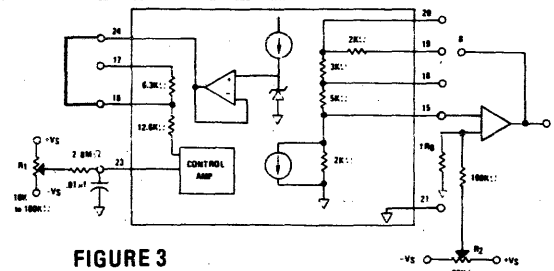


FIGURE 3

**EXTERNAL AMPLIFIER CONNECTIONS**

To use the HI-5680I with an external amplifier, connect as follows:

Range	Pin 17 to	Pin 18 to	Pin 19 to	Pin 20 to
0 to +10V	N.C.	B	18*	19*
0 to +5V	N.C.	B	15	N.C.
$\pm 10V$	15	N.C.	B	N.C.
$\pm 5V$	15	B	18*	19*
$\pm 2.5V$	15	B	15	N.C.

\*these connections help reduce stray capacitance in the feedback loop.  
† $R_B$  should equal the DAC's output resistance, which is  $2K\Omega // R_{FEEDBACK}$ .

**GAIN AND OFFSET CALIBRATION**

(Applies to Figure 2 and 3.)

UNIPOLAR CALIBRATION	
Step 1: Offset	Turn all bits OFF (11...1) Adjust $R_2$ for zero volts out
Step 2: Gain	Turn all bits ON (00...0) Adjust $R_1$ for FS-1LSB That is: 4.9988 for 0 to +5V range 9.9976 for 0 to +10V range
BIPOLAR CALIBRATION	
Step 1: Offset	Turn all bits OFF (11...1) Adjust $R_2$ for Negative FS That is: -10V for $\pm 10V$ range -5V for $\pm 5V$ range -2.5V for $\pm 2.5V$ range
Step 2: Gain	Turn all bits ON (00...0) Adjust $R_1$ for positive FS-1LSB That is: +9.9951V for $\pm 10V$ range +4.9976V for $\pm 5V$ range +2.4988V for $\pm 2.5V$ range
This Bipolar procedure adjusts the output range end points. The maximum error at zero (half scale) will not exceed the Linearity error. See the "Accuracy" specifications.	



# HI-5685

## High Performance Monolithic 12 Bit Digital-to-Analog Converter

### FEATURES

- DAC 85 ALTERNATE SOURCE
- MONOLITHIC CONSTRUCTION (SINGLE CHIP)
- FAST SETTLING
- GUARANTEED MONOTONIC -25°C TO +85°C
- WAFER LASER TRIMMED
- APPLICATIONS RESISTORS ON-CHIP
- ON-BOARD REFERENCE
- DIELECTRIC ISOLATION (DI) PROCESSING
- ±12V POWER SUPPLY OPERATION

### APPLICATIONS

- HIGH SPEED A/D CONVERTERS
- PRECISION INSTRUMENTATION
- CRT DISPLAY GENERATION

### DESCRIPTION

The HI-5685 is a monolithic direct replacement for the popular DAC85-CBI and the ADDAC85LD-CBI. Single chip construction along with several design innovations make the HI-5685 the optimum choice for low cost, high reliability applications.

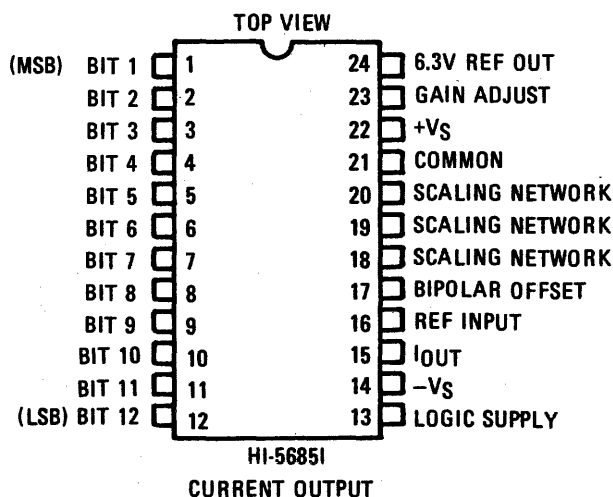
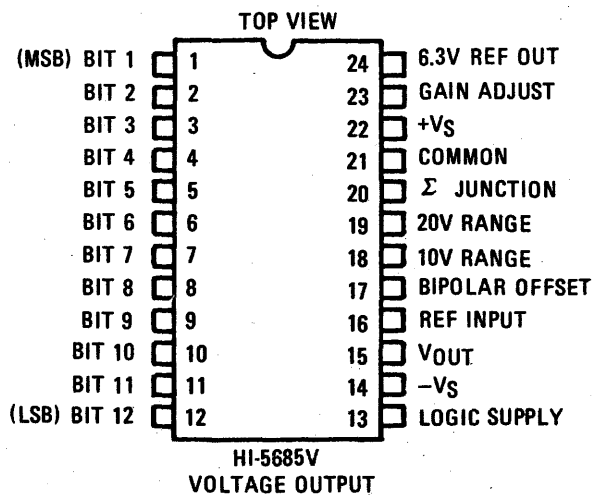
Harris' unique Dielectric Isolation (DI) processing reduces internal parasitics resulting in fast switching times and minimum glitch. On board span resistors are provided for good tracking over temperature, and are laser trimmed to high accuracy. These may be used with the on-board op-amp (voltage output models; HI-5685V), or with a user supplied external amplifier (HI-5685I).

Internally, the HI-5685 eliminates code dependent ground currents by routing current from the positive supply to the internal ground node, as determined by an auxiliary R-2R ladder. This results in a cancellation of code dependent ground currents allowing virtually zero variation in current through the package common, pin 21.

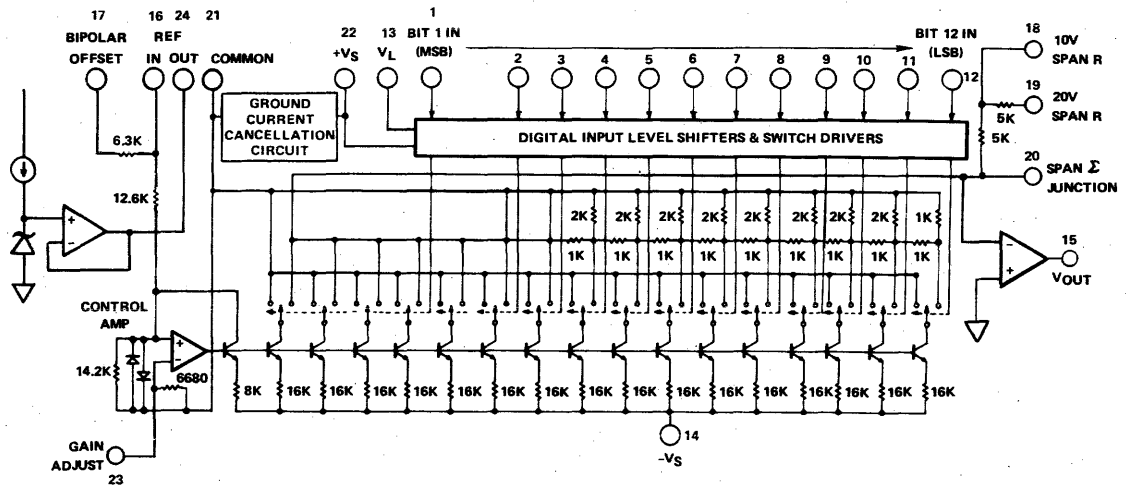
The HI-5685 and HI-5685A are available in both current and voltage output models which are guaranteed over the -25°C to +85°C temperature range. All models include a buried zener reference featuring low temperature coefficient. In addition, the voltage output models include an on-board output amplifier. Both versions operate with a +5V logic supply and a ±V<sub>S</sub> in the range of ±(11.4V to 16.5V).

The HI-5685A offers exceptionally low drift over temperature. Gain drift is a maximum ±10ppm/°C, over -25°C to +85°C.

### PINOUTS

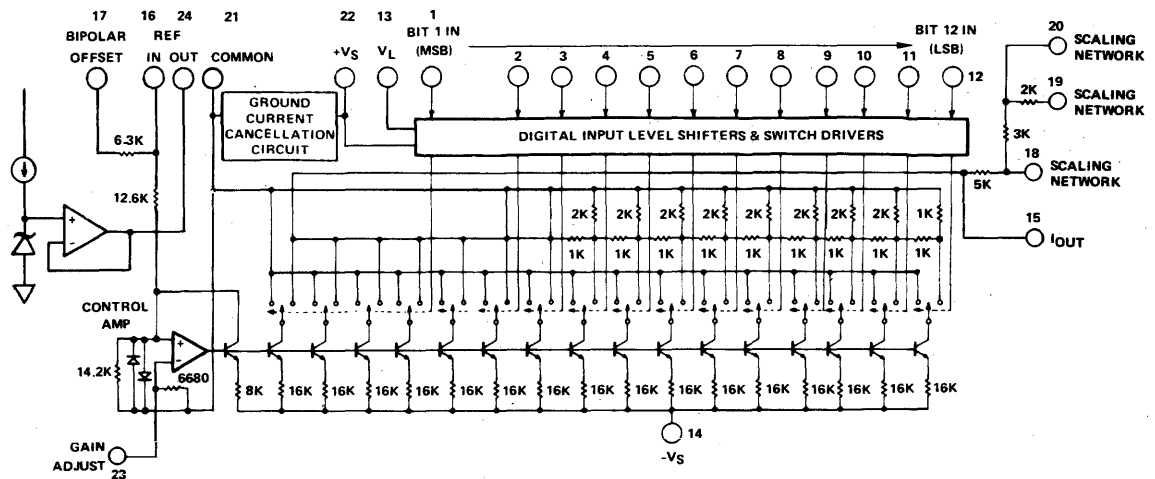


**FUNCTIONAL DIAGRAM VOLTAGE OUTPUT**



HI-5685 V

**FUNCTIONAL DIAGRAM CURRENT OUTPUT**



HI-5685 I

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (1)

Power Supply Inputs	+V <sub>S</sub>	+20V	Power Dissipation	Pd, Package	1000mW
	-V <sub>S</sub>	-20V			
	+V <sub>LOGIC</sub>	+20V	Operating Temperature Range	HI-5685I/V-4	-25°C to +85°C
Reference	Input (pin 16)	±V <sub>S</sub>		HI-5685AI/V-4	-25°C to +85°C
	Output drain	2.5mA			
Digital Inputs	Bits 1 to 12	-1V to +12V	Storage Temperature Range		-65° to +150°C

## ELECTRICAL SPECIFICATIONS

(T<sub>A</sub> = +25°C, V<sub>S</sub> = ±15V, V<sub>LOGIC</sub> = 5V, PIN 16 CONNECTED TO PIN 24 UNLESS OTHERWISE SPECIFIED)

PARAMETER	CONDITIONS	HI-5685			
		MIN	TYP	MAX	UNITS
DIGITAL INPUT (3)					
Resolution				12	Bits
Logic Levels	TTL Compatible				
Logic "1"	at +1 μA	+2		+5.5	V
Logic "0"	at -100 μA	0		+0.8	V
Accuracy (3)					
Linearity Error	at +25°C -25°C to +85°C			±½ ±½	LSB LSB
Differential Lin. Error			±½		LSB
Gain Error (2)			±0.1		%FSR (4)
Offset Error (2)			±0.05		%FSR
Monotonicity	-25°C to +85°C		GUARANTEED		
DRIFT (3) HI-5685	-25°C to +85°C				
Gain				±20	
Offset					
Unipolar			±1	±3	PPM/°C
Bipolar			±5	±10	
DRIFT (3) HI-5685A (Low Drift)	-25°C to +85°C				
Gain				±10	
Offset					
Unipolar			±1		PPM/°C
Bipolar				±5	
CONVERSION SPEED					
Voltage Models					
Settling Time (3)	to ±0.01% of FSR for FSR Change				
With 10KΩ Feedback			3		μs
With 5KΩ Feedback			1.5		μs
For 1 LSB Change			1.5		μs
Slew Rate			15		V/μs
Current Models					
Settling Time (3)	to ±0.01% of FSR for FSR Change				
10 to 100Ω load			300		ns
1KΩ load			1.0		μs





PARAMETER	CONDITIONS	HI-5685			
		MIN	TYP	MAX	UNITS
<b>ANALOG OUTPUT</b>	Full Scale				
Voltage Models					
Output Current		±5			mA
Output Impedance (DC)			0.05		Ω
Current Models					
Output Current					
Unipolar		-1.6	-2	-2.4	mA
Bipolar		±0.8	±1	±1.2	mA
Output Resistance					
Unipolar		2.0		KΩ	
Bipolar		2.0		KΩ	
Compliance (3)		-2.5		+10	V
<b>INTERNAL REFERENCE</b>					
Output voltage		+6.174	+6.3	+6.426	V
Output Impedance			1.5		Ω
External Current				+2.5	mA
Tempco of Drift			±10	±20	PPM/°C
<b>POWER SUPPLY SENSITIVITY (3)</b>					
+15V				.002	$\frac{\%FSR}{\Delta V_s}$
-15V				.002	
+5V				.002	
<b>POWER SUPPLY REQUIREMENTS(5)</b>					
Range					
+15V		+11.4	+15	+16.5	V
-15V		-11.4	-15	-16.5	V
+5V		+4.5	+5	+16.5	V
Current					
+15V			8	11	mA
-15V			-12	-20	mA
+5V			4.5	8	mA

NOTES:

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. Adjustable to zero using external potentiometers.
3. See Definitions.
4. FSR is "full scale range" and is 20V for ±10V range, 10V for ±5V range, etc., or 2mA (±20%) for current output.
5. The HI-5685 will operate with supply voltages as low as ±11.4V. It is recommended that output voltage range -10V to +10V not be used if the supply voltages are less than ±12.5V.



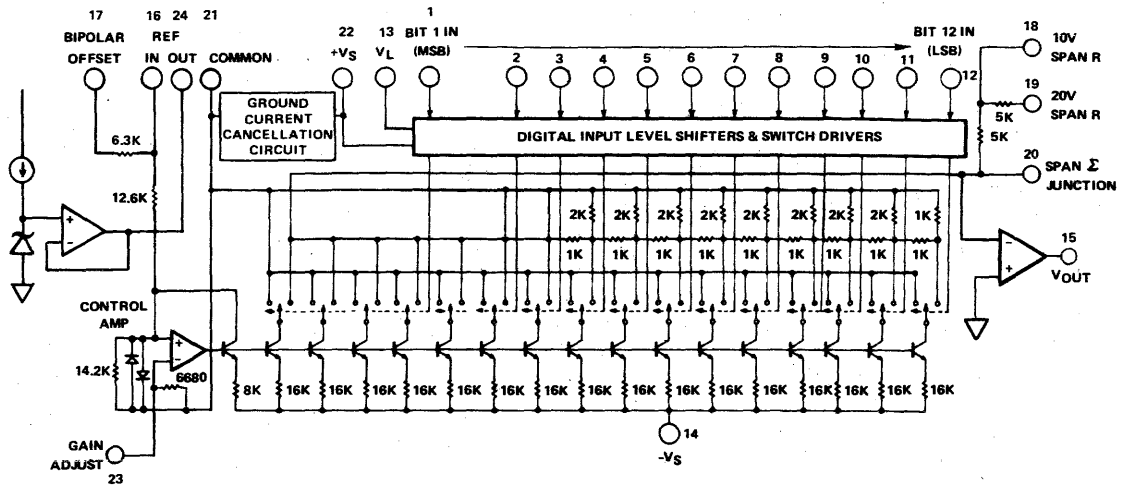
# HI-5687

*Wide Temperature Range  
Monolithic 12 Bit  
Digital-to-Analog Converter*

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• DAC 87 ALTERNATE SOURCE</li> <li>• MONOLITHIC CONSTRUCTION (SINGLE CHIP)</li> <li>• FAST SETTling</li> <li>• GUARANTEED SPECIFICATIONS      -55°C to 125°C</li> <li>• WAFER LASER TRIMMED</li> <li>• APPLICATIONS RESISTORS ON-CHIP</li> <li>• ON-BOARD REFERENCE</li> <li>• DIELECTRIC ISOLATION (DI) PROCESSING</li> <li>• ±12V POWER SUPPLY OPERATION</li> <li>• MIL STD 883 PROCESSING AVAILABLE</li> </ul>	<p>The HI-5687 is a monolithic direct replacement for the popular DAC87-CBI wide temperature range d-to-a converter. Single chip construction, along with several design innovations make the HI-5687 the optimum choice for low cost, high reliability applications.</p> <p>Harris' unique Dielectric Isolation (DI) processing reduces internal parasitics resulting in fast switching times and minimum glitch. On board span resistors are provided for good tracking over temperature, and are laser trimmed to high accuracy. These may be used with the on-board op-amp (voltage output models; HI-5687V), or with a user supplied external amplifier (HI-5687I).</p> <p>Internally, the HI-5687 eliminates code dependent ground currents by routing current from the positive supply to the internal ground mode, as determined by an auxiliary R-2R ladder. This results in a cancellation of code dependent ground currents allowing virtually zero variation in current through the package common, pin 21.</p>
<p><b>APPLICATIONS</b></p> <ul style="list-style-type: none"> <li>• HIGH SPEED A/D CONVERTERS</li> <li>• PRECISION INSTRUMENTATION</li> <li>• CRT DISPLAY GENERATION</li> </ul>	<p>The HI-5687 is available in both current and voltage output models which are 100% tested over the -55°C to +125°C temperature range. All models include a buried zener reference featuring low temperature coefficient. In addition, the voltage output models include an on-board output amplifier. Both versions operate with a +5V logic supply and a <math>\pm V_S</math> in the range of <math>\pm(11.4V</math> to <math>16.5V)</math>.</p> <p>Processing to MIL-STD-883A CLASS B is available. See Ordering Information.</p>

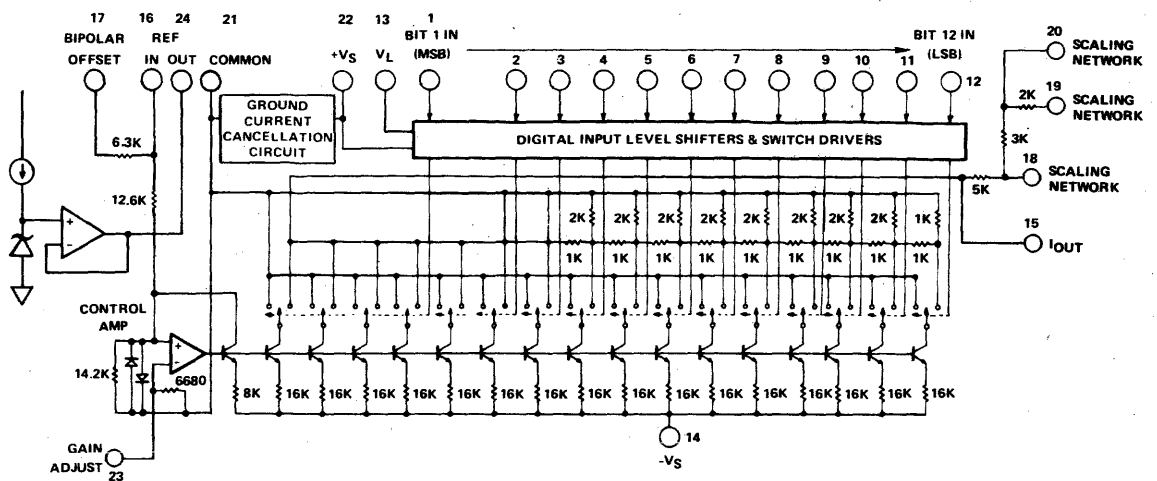
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# FUNCTIONAL DIAGRAM VOLTAGE OUTPUT



HI-5687 V

# FUNCTIONAL DIAGRAM CURRENT OUTPUT



HI-5687 I

# SPECIFICATIONS



INTERFACE

Harris Semiconductor

## ABSOLUTE MAXIMUM RATINGS <sup>(1)</sup>

Power Supply Inputs	+V <sub>S</sub> +20V -V <sub>S</sub> -20V +V <sub>LOGIC</sub> +20V	Power Dissipation Pd, Package	1000mW
Reference	Input (pin 16) ± V <sub>S</sub> Output drain 2.5mA	Operating Temperature Range	HI-5687I/V-2 -55°C to +125°C HI-5687I/V-8 -55°C to +125°C
Digital Inputs	Bits 1 to 12 -1V to +12V	Storage Temperature Range	-65°C to +150°C

## ELECTRICAL SPECIFICATIONS

(T<sub>A</sub> = +25°C, V<sub>S</sub> = ±15V, V<sub>LOGIC</sub> = +5V, PIN 16 CONNECTED TO PIN 24 UNLESS OTHERWISE SPECIFIED.)

PARAMETER	CONDITIONS	HI-5687			
		MIN	TYP	MAX	UNITS
<b>DIGITAL INPUT (3)</b>					
Resolution				12	Bits
Logic Levels	TTL Compatible				
Logic "1"	at +1μA	+2		+5.5	V
Logic "0"	at -100μA	0		+0.8	V
<b>ACCURACY (3)</b>					
Linearity Error	At +25°C -55°C to +125°C		±½	±½ +¾	LSB LSB
Differential Lin. Error	at +25°C -55°C to +125°C		±½	±¾ ±1	LSB LSB (4)
Gain Error (2)			±0.1	±0.2	%FSR
Offset Error (2)			±0.05	±0.1	%FSR
Monotonicity	-55°C to +125°C		GUARANTEED		
<b>DRIFT (3)</b>	-55°C to +125°C				
Total Bipolar Drift (includes gain, offset and linearity drifts)			±15	±30	PPM/°C
Total Error					
Unipolar			±0.13	±0.3	%FSR
Bipolar			±0.12	±0.24	%FSR
Gain					
including internal reference			±10	±25	PPM/°C
excluding internal reference			±5	±10	PPM/°C
Unipolar Offset			±1	±3	PPM/°C
Bipolar Offset			±5	±10	PPM/°C
<b>CONVERSION SPEED</b>					
Voltage Models					
Settling Time (3)	to ±0.01% of FSR for FSR Change				
With 10KΩ Feedback			3		μs
With 5KΩ Feedback			1.5		μs
For 1 LSB Change			1.5		μs
Slew Rate			15		V/μs
Current Models					
Settling Time (3)	to ±0.01% of FSR for FSR Change				
10 to 100Ω load			300		ns
1KΩ load			1.0		μs



PARAMETER	CONDITIONS	HI-5687			
		MIN	TYP	MAX	UNITS
<b>ANALOG OUTPUT</b>	Full Scale				
Voltage Models					
Output Current		+5			mA
Output Impedance (DC)			0.05		$\Omega$
Current Models					
Output Current					
Unipolar		-1.6	-2	-2.4	mA
Bipolar	$\pm 0.8$	$\pm 1$	$\pm 1.2$	mA	
Output Impedance					
Unipolar		2.0		$K\Omega$	
Bipolar		2.0		$K\Omega$	
Compliance (3)		-1.5		+ 10.5	V
<b>INTERNAL REFERENCE</b>					
Output Voltage		+6.174	+6.3	+6.426	V
Output Impedance			1.5		$\Omega$
External Current			$\pm 5$	+2.5	mA
Tempco of Drift				$\pm 10$	PPM/ $^{\circ}C$
<b>POWER SUPPLY SENSITIVITY (3)</b>					
+15V				$\pm 0.002$	$\frac{\%FSR}{\Delta V_s}$
-15V				$\pm 0.002$	
+5V				$\pm 0.002$	
<b>POWER SUPPLY REQUIREMENTS (5)</b>					
Range					
+15V		+11.4	+15	+16.5	V
-15V		-11.4	-15	-16.5	V
+5V		+4.5	+5	+16.5	V
Current					
+15V			8	11	mA
-15V			-12	-20	mA
+5V			4.5	8	mA

**NOTES:**

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. Adjustable to zero using external potentiometers.
3. See Definitions.
4. FSR is a "full scale range" and is 20V for  $\pm 10V$  range, 10V for  $\pm 5V$  range, etc., or 2mA ( $\pm 20\%$ ) for current output.
5. The HI-5687 will operate with supply voltages as low as  $\pm 11.4V$ . It is recommended that output voltage ranges  $-10V$  to  $+10V$  and not be used if the supply voltages are less than  $\pm 12.5V$ .



# HI-DAC16B/C

## 16-Bit D to A Converter

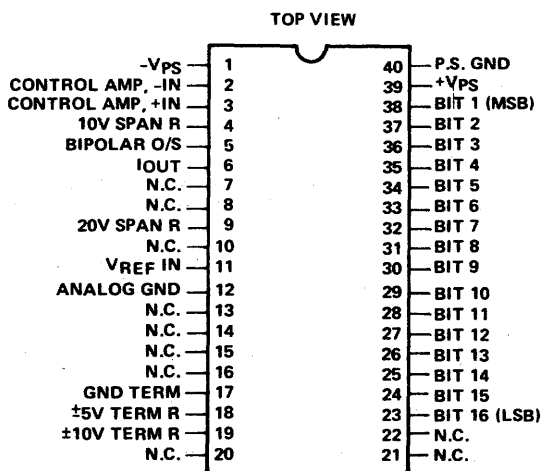
### FEATURES

- 16 BIT RESOLUTION
- MONOLITHIC DI BIPOLAR CONSTRUCTION
- FAST SETTLING TIME  $1\mu\text{s TO } .003\%FS$
- LOW DIFF. NONLIN. DRIFT  $\pm 0.3\text{ppm}/^\circ\text{C}$
- LOW GAIN DRIFT  $\pm 1\text{ppm}/^\circ\text{C}$
- ON-CHIP SPAN & OFFSET RESISTORS
- TTL/5V-CMOS COMPATIBLE
- LOW UNIPOLAR OFFSET  $\leq 1/2\text{LSB}@ +25^\circ\text{C}$
- LOW UNIPOLAR OFFSET T.C.  $\pm 0.2\text{ppm}/^\circ\text{C}$
- EXCELLENT STABILITY

### APPLICATIONS

- HIGH RESOLUTION CONTROL SYSTEMS
- HIGH FIDELITY AUDIO RECONSTRUCTION
- PRECISION FUNCTION GENERATION AND INSTRUMENTATION

### PINOUT



### DESCRIPTION

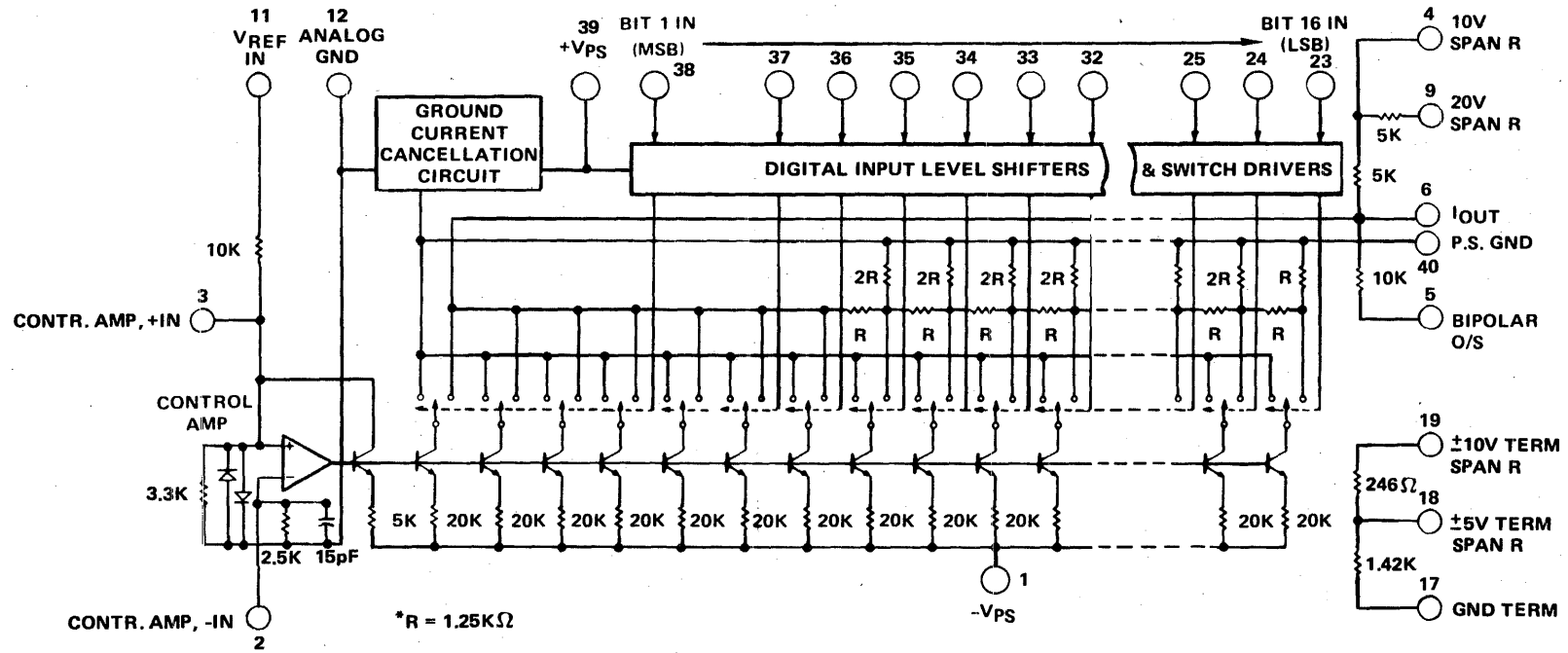
The HARRIS HI-DAC16 is a 16-bit, current output D/A converter. Single chip construction includes thin-film application resistors for use with an external op amp. These permit standard output voltage ranges of 0 to +5V, 0 to +10V,  $\pm 2.5V$ ,  $\pm 5V$  and  $\pm 10V$ .

Reference and span resistors have adjacent placement on the chip for optimum match and thermal tracking. Furthermore, this layout feature helps minimize the superposition error caused by self-heating of the span resistor, reducing it to less than 1/10LSB. This and other design innovations have produced exceptionally stable operation over temperature. Typical temperature coefficients are  $\pm 1\text{ppm}/^\circ\text{C}$  for gain error and  $0.3\text{ppm}/^\circ\text{C}$  for differential non-linearity error.

The internal architecture is an extension of the earlier HI-562 with several major improvements. All code dependent ground currents are steered to a separate non-critical path, namely, power supply ground. This feature allows the precision ground of the converter to be sensed with virtually zero voltage drop referred to system ground. The result is the complete elimination of non-linearities due to code dependent ground currents while yielding an extremely low unipolar offset of less than 1/2LSB. Because of this separation, the user may route the precision ground some distance to the system ground without degrading converter accuracy.

The HARRIS HI-DAC16 delivers a stable, accurate output without sacrifice in speed. Settling time to within  $\pm 0.003\%$  is one microsecond. Overall performance of this monolithic device should be attractive for applications such as high fidelity audio and high-resolution control systems.

Typical power requirement is 450 MW, from the +15V and -15V supplies combined. The package is a 40 pin ceramic DIP. Two accuracy grades are offered.





**ABSOLUTE MAXIMUM RATINGS** (Referred to Ground)

Power Supply Inputs	$V_{ps+}$	+20V	Power Dissipation $P_d$ , Package	1000 mW
	$V_{ps-}$	-20V	Operating Temperature Range	
Reference Inputs	$V_{REF}$ (Hi)	$\pm V_{ps}$		
Digital Inputs	Bits 1 to 16	-1V, +12V	HI-DAC 16B/C	0°C to +75°C
Outputs		$\pm V_{ps}$	Storage Temperature Range	-65°C to +150°C

**ELECTRICAL CHARACTERISTICS** ( $T_A = +25^\circ\text{C}$ ,  $V_{ps} = \pm 15\text{V}$ ,  $V_{ref} = +10\text{V}$ , unless otherwise specified)

PARAMETER	CONDITIONS	HI-DAC 16B			HI-DAC 16C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Digital Inputs	Bit ON "Logic 1" Bit OFF "Logic 0"							
Input Voltage Logic "1" Logic "0"	Over full temp. range	2.0		0.8	2.0		0.8	V
Input Current Logic "1" Logic "0"			-50			20		500
Reference Input Input Resistance Input Voltage			10			10		k $\Omega$ V

**TRANSFER CHARACTERISTICS**

Resolution	Over full temp. range		16			16		Bits
Nonlinearity	@+25°C Over full temp. range		$\pm 0.023$	$\pm 0.045$		$\pm 0.045$	$\pm 0.09$	%FSR(3)
Differential Nonlinearity	@+25°C Over full temp. range		$\pm 0.015$	$\pm 0.03$		$\pm 0.03$	$\pm 0.06$	%FSR
Relative Accuracy (5)	With 100 $\Omega$ (1%) Trim Resistors							
Unipolar Gain Error	All Bits ON		$\pm 1$	$\pm 25$		$\pm 1$	$\pm 25$	%FSR
Bipolar Offset Error	All Bits OFF		$\pm 15$			$\pm 15$	$\pm 42$	%FSR
Unipolar Offset Error			$\pm 0.02$			$\pm 0.02$	$\pm 05$	
Adjustment Range	See Operating Instructions							
Gain	Using trim potentiometers as shown in Figure 1			$\pm 3$			$\pm 3$	%FSR
Bipolar Offset				$\pm 43$			$\pm 43$	
Temperature Stability	Drift specified with internal span resistors for voltage output							
Gain Drift (2)	Over full temp. range		$\pm 1$	$\pm 5$		$\pm 1$	$\pm 5$	ppm of FSR/°C
Offset Drift (2)			$\pm 2$			$\pm 2$		
Unipolar Offset	All Bits OFF		$\pm 5$			$\pm 5$		
Bipolar Offset	Over full temp. range							
Differential Nonlinearity			$\pm 0.3$			$\pm 0.3$		
Settling Time (2) to $\pm 0.03\%$ FS	All Bits ON-to-OFF or OFF-to-ON		1.0			1.0		$\mu\text{s}$





PARAMETER	CONDITIONS	HI-DAC 16B			HI-DAC 16C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
Glitch (2)	From 011 ... 1 to 100 ... 0 or 100 ... 0 to 011 ... 1		1300			1300		mV-ns
Power Supply (2) Rejection Ratio, PSRR (3)  $V_{ps+}$ $V_{ps-}$			1.5 1.5			1.5 1.5		ppm of FSR/% $V_{ps}$

**OUTPUT CHARACTERISTICS**

Output Current Unipolar Bipolar		-1.6 $\pm 0.8$	-2 $\pm 1$	-2.4 $\pm 1.2$	-1.6 $\pm 0.8$	-2 $\pm 1$	-2.4 $\pm 1.2$	mA
Resistance			2.5K			2.5K		ohms
Capacitance			10			10		pF
Output Voltage Ranges Unipolar Bipolar	Using external op amp and internal scaling resistors. See Figure 1 and Table 1 for connections		0 to +5 0 to +10 $\pm 2.5$ $\pm 5$ $\pm 10$			0 to +5 0 to +10 $\pm 2.5$ $\pm 5$ $\pm 10$		V
Compliance Limit (2)		-3		+10	-3		+10	V
Compliance Voltage (2)	Over full temp. range		$\pm 1$			$\pm 1$		V
Output Noise	0.1 to 5MHz (All Bits ON)		30			30		$\mu$ VRMS

**POWER REQUIREMENTS**

$V_{ps+}$ (7) $V_{ps-}$	Over full temp. range	13.5 -13.5	+15 -15	16.5 -16.5	13.5 -13.5	+15 -15	16.5 -16.5	V
$I_{ps+}$ (4) $I_{ps-}$ (4)	All Bits ON or OFF full temp. range	-25	+13 -18	+18	-25	+13 -18	+18	mA
Power Dissipation			465			465		mW

**NOTES:**

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. See Definitions.
3. FSR is "full scale range" and is 20V for  $\pm 10$ V range, 10V for  $\pm 5$ V range, etc., or 2mA ( $\pm 20\%$ ) for current output.
4. After 30 seconds warm-up.
5. Using an external op amp with internal span resistors and specified external trim resistors in place of potentiometers R<sub>1</sub> and R<sub>2</sub>. Errors are adjustable to zero using R<sub>1</sub> and R<sub>2</sub> potentiometers. (See Operating Instructions Figure 2.)



# HI-5660

## High Speed Monolithic Digital-to-Analog Converter

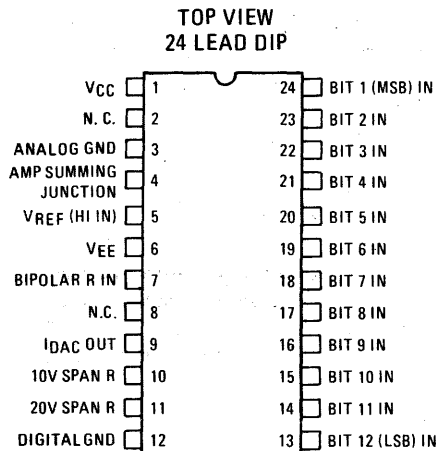
### FEATURES

- MONOLITHIC CONSTRUCTION
- FAST SETTLING (TO  $\pm 1/2$  LSB) 350ns
- $\pm 1/2$  LSB MAX. NONLINEARITY GUARANTEED OVER TEMPERATURE
- INTERNAL CANCELLATION OF GROUND CURRENT
- EXCELLENT POWER SUPPLY REJECTION 1ppm/%PS
- LOW COST

### APPLICATIONS

- HIGH SPEED A/D CONVERTERS
- CRT DISPLAYS
- WAVEFORM SYNTHESIS

### PINOUT



### DESCRIPTION

The HI-5660 is a current output, 12 bit monolithic digital-to-analog converter. It offers high speed plus enhanced accuracy, through internal cancellation of ground currents.

Electrical performance is similar to that of the AD566A. Pinouts are identical except for pin 1, which requires a +5V supply (versus no connection on the AD566A).

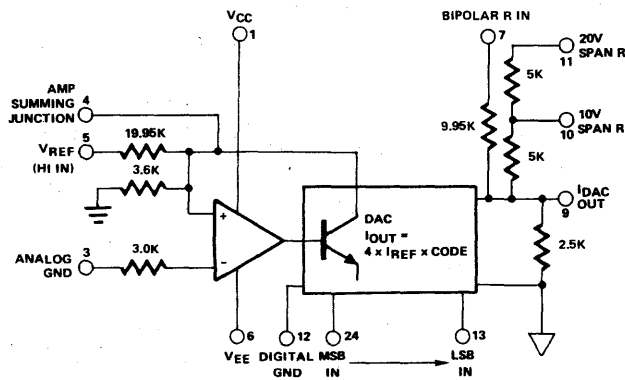
Fabrication of the HI-5660 features the Harris bipolar dielectric isolation process, which eliminates latchup and minimizes parasitic capacitance and leakage currents. The chip includes nichrome thin-film resistors, laser trimmed at the wafer level to a maximum linearity error of  $\pm 1/4$  LSB at +25°C.

Near zero current in the Analog Ground terminal simplifies use of the HI-5660 by minimizing noise and offsets between the package and the system analog ground. This is accomplished by adding a complement current to the internal ground from an auxiliary R-2R ladder, and then supplying the resultant DC current from the positive power supply.

The Harris HA-1608 +10V precision reference is recommended for use with the HI-5660 in non-multiplying applications.

The HI-5660 is offered in two accuracy grades each for the commercial and military temperature ranges. Package is a 24 pin plastic or ceramic DIP, and power requirements are +15V, -15V.

### FUNCTIONAL DIAGRAM





**ABSOLUTE MAXIMUM RATINGS\***

V <sub>CC</sub> to Power Ground	0V to +18V	10V Span R to Reference Ground	±12V
V <sub>EE</sub> to Power Ground	0V to -18V	20V Span R to Reference Ground	±24V
Voltage on DAC Output (Pin 9)	-3V to +12V	Package Power Dissipation	
Digital Inputs (Pins 13-24) to Power Ground	-1V to +7.0V	Ceramic	1000mW
Ref In to Reference Ground	±12V	Plastic	750mW
Bipolar Offset to Reference Ground	±12V		

\*Absolute maximum ratings are limiting values beyond which the serviceability of the circuit may be impaired.

**ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = +25°C, V<sub>CC</sub> = +15V, V<sub>EE</sub> = -15V, Unless Otherwise Specified)

MODEL	HI-5660-5			HI-5660A-5			UNITS
	MIN	TYP	MAX	MIN	TYP	MAX	
<b>DATA INPUTS</b> (Note 1) (Pins 13 to 24)							
TTL or 5V CMOS (T <sub>MIN</sub> to T <sub>MAX</sub> )							
Input Voltage							
Bit ON Logic "1"	2.0		5.5	2.0		5.5	V
Bit OFF Logic "0"	0.0		0.8	0.0		0.8	V
Logic Current (Each Bit)							μA
Bit ON Logic "1"		2	10		2	10	μA
Bit OFF Logic "0"		-10	-50		-10	-50	μA
<b>RESOLUTION</b>							
			12			12	Bits
<b>OUTPUT</b>							
Current							mA
Unipolar (All Bits On)	-1.6	-2.0	-2.4	-1.6	-2.0	-2.4	mA
Bipolar (All Bits on or Off)	±0.8	±1.0	±1.2	±0.8	±1.0	±1.2	mA
Resistance (Exclusive of Span Resistors)	2.0K	2.5K	3.0K	2.0K	2.5K	3.0K	Ω
Offset							% of FS
Unipolar		.01	.05		.01	.05	% of FS
Bipolar (Figure 2, R <sub>3</sub> = 50Ω Fixed)		.05	.15		.05	.15	% of FS
Capacitance		25			25		pF
Compliance Voltage, T <sub>MIN</sub> to T <sub>MAX</sub>	-3		+12	-3		+12	V
<b>ACCURACY</b> (Error Relative to Full Scale)							
							LSB
+25°C		±1/4 (0.006)	±1/2 (0.012)		±1/8 (0.003)	±1/4 (0.006)	% of FS
T <sub>MIN</sub> to T <sub>MAX</sub>		±1/2 (0.012)	±3/4 (0.018)		±1/4 (0.006)	±1/2 (0.012)	% of FS
<b>DIFFERENTIAL NONLINEARITY</b>							
							LSB
+25°C		±1/2	±3/4		±1/4	±1/2	LSB
T <sub>MIN</sub> to T <sub>MAX</sub>	MONOTONICITY GUARANTEED (±1 LSB MAX)						
<b>TEMPERATURE COEFFICIENTS</b>							
							ppm/°C
Unipolar Zero		1	2		1	2	ppm/°C
Bipolar Zero		5	10		5	10	ppm/°C
Gain (Full Scale)		7	10		7	10	ppm/°C
Differential Nonlinearity		2	6		2	2	ppm/°C
<b>SETTLING TIME TO 1/2 LSB</b>							
							ns
With High Z External Load (Note 2)		500			500		ns
With 75Ω External Load		250			250		ns

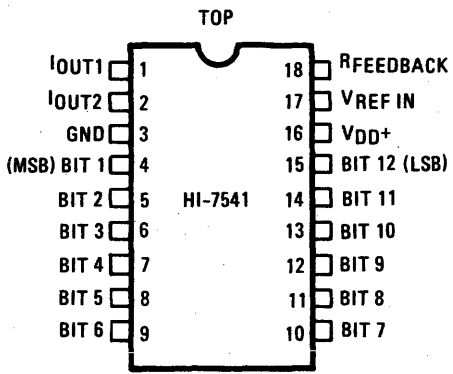
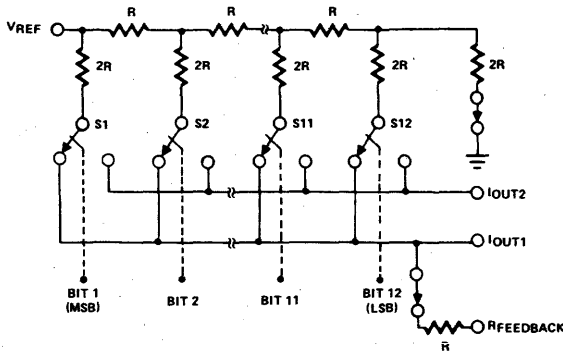


MODEL	HI-5660-5			HI-5660A-5			UNITS
	MIN	TYP	MAX	MIN	TYP	MAX	
<b>TEMPERATURE RANGE</b>							
Operating	0		+75	0		+75	°C
Storage	-25		+150	-25		+150	°C
<b>POWER REQUIREMENTS</b>							
VCC, +4.5V to +16.5VDC		7	12		7	12	mA
VEE, -11.4 to -16.5VDC		-13	-17		-13	-17	mA
<b>POWER SUPPLY GAIN SENSITIVITY</b>							
VCC = +4.5 to +16.5VDC; VEE = -15V		1	10		1	10	ppm of FS/%
VEE = -11.4 to -16.5VDC; VCC = +15V		1	10		1	10	ppm of FS/%
<b>PROGRAMMABLE OUTPUT RANGES (See Table 1)</b>							
		0 to +5			0 to +5		V
		-2.5 to +2.5			-2.5 to +2.5		V
		0 to +10			0 to +10		V
		-5 to +5			-5 to +5		V
		-10 to +10			-10 to +10		V
<b>EXTERNAL ADJUSTMENTS</b>							
Gain Error with Fixed 50Ω Resistor for R <sub>2</sub> (Figure 1)		±0.1	±0.25		±0.1	±0.25	% of FS
Bipolar Offset Error with Fixed 50Ω Resistor for R <sub>3</sub> (Figure 2)		±0.05	±0.15		±0.05	±0.1	% of FS
Gain Adjustment Range (Figure 1)	±0.25			±0.25			% of FS
Bipolar Offset Adjustment Range (Fig. 2)	±0.15			±0.15			% of FS
<b>REFERENCE INPUT</b>							
Input Impedance	16K	20K	24K	16K	20K	24K	Ω
<b>POWER DISSIPATION</b>							
		230	330		230	330	mW
<b>MULTIPLYING MODE PERFORMANCE (All Models)</b>							
Quadrants	Two (2): Bipolar Operation at Digital Input Only.						
Reference Voltage	Unipolar: +10V Max, +2V Min.						
Accuracy	10 Bits (±0.05% of Reduced F.S.) for 2V <sub>DC</sub> Reference Voltage.						
Reference Feedthrough (Unipolar Mode, All Bits OFF, and +2V to +10V (p-p), Sinewave Frequency for 1/2 LSB (p-p) Feedthrough)	22kHz Typical						
Output Slew Rate 10%-90%	1.3mA/μs						
90%-10%	1.3mA/μs						
Output Settling Time (All Bits ON and a +2V to +10V Step Change in Reference Voltage)	1.5μs to 0.01% F.S.						
<b>CONTROL AMPLIFIER</b>							
Full Power Bandwidth (+10V to +3V)		200			200		kHz
Small Signal Closed-Loop Bandwidth		2.4			2.4		MHz

**NOTES:**

1. The Digital Input Levels are Guaranteed but not Tested Over the Temperature Range.
2. See Settling Time Section.

## 12 Bit Multiplying Monolithic Digital-to- Analog Converter

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• FULL FOUR QUADRANT MULTIPLICATION</li> <li>• .01% RELATIVE ACCURACY OVER TEMPERATURE</li> <li>• LOW OUTPUT CAPACITANCE <span style="float: right;">100pF MAX</span></li> <li>• TTL/CMOS COMPATIBLE</li> <li>• MONOLITHIC CONSTRUCTION</li> <li>• VERY LOW OUTPUT LEAKAGE CURRENT <span style="float: right;"><math>\pm 100\text{nA MAX}</math></span></li> <li>• LOW-GAIN ERROR <span style="float: right;">0.1%</span></li> </ul>	<p>The Harris HI-7541 is a 12-Bit Monolithic Digital to Analog converter, offering full four quadrant multiplying capability. The chip features dielectrically isolated CMOS technology to assure fast settling time and freedom from latch-up. Included are thin film ladder and applications resistors, laser trimmed for accuracy over the full operating temperature range.</p> <p>The HI-7541 is recommended as a high performance direct replacement for the AD7541 device. It operates on a single +15V supply and is available in an 18-pin ceramic package as well as in dice form. Screening to MIL-STD-883 method 5004 class B is available.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• PROGRAMMABLE GAIN AMPLIFIERS</li> <li>• PROGRAMMABLE FUNCTION GENERATION</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<div style="text-align: center;">TOP</div> 	 <p>DIGITAL INPUTS (DTL, TTL, CMOS COMPATIBLE) LOGIC: A SWITCH IS CLOSED TO IOUT1 FOR ITS DIGITAL INPUT IN A HIGH (LOGIC 1) STATE.</p>



# SPECIFICATIONS

INTERFACE

Harris Semiconductor

## ABSOLUTE MAXIMUM RATINGS (Referred to Ground)1

Power Supply Inputs V <sub>DD</sub>	+17V	Power Dissipation (Package) up to +75°C	450mW
Reference Inputs V <sub>REF</sub> (Hi)	±25V	Derate above +75°C by 6mW/°C.	
Digital Input Range Bits 1-12	V <sub>DD</sub> to GND	Operating Temperature Range	
		HI-7541SD/TD/SO	-55°C to +125°C
		HI-7541AD/BD	-25°C to +85°C
		HI-7541JN/KN/JO	0°C to +75°C
Output Voltage (Pins 1 and 2)	-400mV to V <sub>DD</sub>	HI-7541SD/883 AND TD/883.	-55°C to +125°C
		Storage Temperature Range	-65°C to +150°C

## ELECTRICAL CHARACTERISTICS (@25°C, V<sub>DD</sub> = +15V, V<sub>REF</sub> = +10V Unless otherwise noted)

PARAMETER	CONDITIONS	HI-7541KN/BD/TD			HI-7541JN/AD/SD/JO/SO			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

### INPUT CHARACTERISTICS

Digital Inputs	Bit ON = "Logic 1" Bit OFF = "Logic 0"	HI-7541KN/BD/TD			HI-7541JN/AD/SD/JO/SO			UNITS
Input Voltage								
Logic 1, V <sub>AH</sub>		2.4			2.4		V	
Logic 0, V <sub>AL</sub>				0.8			0.8 V	
Input Current								
Logic 1	V <sub>IN</sub> =15V			1			1 μA	
Logic 0	V <sub>IN</sub> =0V			-1			-1 μA	
Reference Input								
Input Resistance		7	9	12	7	9	12 KΩ	
Input Voltage		-10		+10	-10		+10 V	

### TRANSFER CHARACTERISTICS

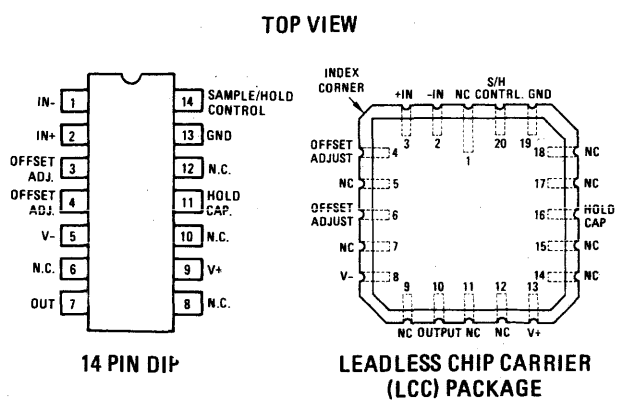
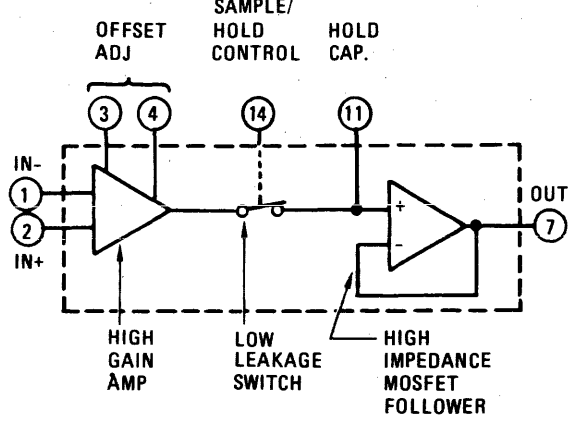
Resolution	Over Full Temp. Range	12			12			Bits
Integral (2)	@+25°C							
Nonlinearity	Over Full Temp. Range			±0.1			±0.2	%FSR
Differential (2)	@+25°C							
Nonlinearity	Over Full Temp. Range			±0.1			±0.2	%FSR
Gain Error (2)	@+25°C			±0.1			±0.2	%FSR
	Over Full Temp. Range			±0.15			±0.25	%FSR
Gain Tempco (2)(5)	Over Full Temp. Range			±5			±5	PPM/°C
Settling Time (2) (5)								
to ±1/2 LSB				1			1	μs
PSRR (2)	14.5V ≤ V <sub>DD</sub> ≤ 15.5V; 25°C			±0.1			±0.1	%FSR/
	Over Full Temp. Range			±0.2			±0.2	%ΔV <sub>DD</sub>

### OUTPUT CHARACTERISTICS

Output (2)	V <sub>REF</sub> = ±10V			±50			±50	nA
Leakage Current	@+25°C							
	Over Full Temp. Range			±100			±100	nA
Capacitance (2) (5)				100			100	pF
Feed Through (2)(5)	V <sub>REF</sub> = 20 V <sub>pp</sub> @ 10kHz			±1			±1	mV <sub>pp</sub>

### POWER REQUIREMENTS

V <sub>DD</sub>	( See Fig. 6, 8, & 9 )	+5	+15	+16	+5	+15	+16	V
I <sub>DD</sub> (3)				2			2	mA

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>● LOW DROOP RATE (<math>C_H = 1000\text{pF}</math>) <math>5\mu\text{V/ms}</math></li> <li>● FAST ACQUISITION TIME (10V STEP TO .01%) <math>5\mu\text{s}</math></li> <li>● HIGH SLEW RATE <math>7\text{V}/\mu\text{s}</math></li> <li>● BANDWIDTH <math>2.5\text{MHz}</math></li> <li>● LOW EFFECTIVE APERTURE DELAY TIME <math>30\text{ns}</math></li> <li>● TTL COMPATIBLE CONTROL INPUT</li> </ul>	<p>The HA-2420/2425 is a monolithic circuit consisting of a high performance operational amplifier with its output in series with an ultra-low leakage analog switch and a MOSFET input unity gain amplifier.</p> <p>With an external holding capacitor connected to the switch output, a versatile, high performance sample-and-hold or track-and-hold circuit is formed. When the switch is closed, the device behaves as an operational amplifier, and any of the standard op amp feedback networks may be connected around the device to control gain, frequency response, etc. When the switch is opened the output will remain at its last level.</p>
APPLICATIONS	<p>Performance as a sample-and-hold compares very favorably with other monolithic, hybrid, modular, and discrete circuits. Accuracy to better than 0.01% is achievable over the temperature range. Fast acquisition is coupled with superior droop characteristics, even at high temperatures. High slew rate, wide bandwidth, and low acquisition time produce excellent dynamic characteristics. The ability to operate at gains greater than 1 frequently eliminates the need for external scaling amplifiers.</p> <p>The device may also be used as a versatile operational amplifier with a gated output for applications such as analog switches, peak holding circuits, etc.</p>
<ul style="list-style-type: none"> <li>● A TO D CONVERSION SYSTEMS</li> <li>● D TO A DEGLITCHER</li> <li>● AUTO ZERO SYSTEMS</li> <li>● PEAK DETECTOR</li> <li>● GATED OP AMP</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;"><b>TOP VIEW</b></p>  <p style="text-align: center;"><b>14 PIN DIP</b></p> <p style="text-align: center;"><b>LEADLESS CHIP CARRIER (LCC) PACKAGE</b></p>	<p style="text-align: center;"><b>FUNCTIONAL DIAGRAM</b></p>  <p style="text-align: center;"><b>HIGH GAIN AMP</b></p> <p style="text-align: center;"><b>LOW LEAKAGE SWITCH</b></p> <p style="text-align: center;"><b>HIGH IMPEDANCE MOSFET FOLLOWER</b></p>



**ABSOLUTE MAXIMUM RATINGS**

Voltage Between V+ and V- Terminals	40V	Operating Temperature Range	
Differential Input Voltage	±24V	HA-2420-2/8	-55°C ≤ TA ≤ +125°C
Digital Input Voltage (Pin 14)	+8V, -15V	HA-2425-5	0°C ≤ TA ≤ +75°C
Output Current	Short Circuit Protected	Storage Temperature Range	-65°C ≤ TA ≤ +150°C
Internal Power Dissipation	300mW (Note 7)		

**ELECTRICAL CHARACTERISTICS** Test Conditions (Unless otherwise specified) V<sub>SUPPLY</sub> = ±15.0V; C<sub>H</sub> = 1000pF; Digital Input (Pin 14), V<sub>IL</sub> = +0.8V (Sample), V<sub>IH</sub> = +2.0V (Hold)

PARAMETER	TEMP	HA-2420-2			HA-2425-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
*Offset Voltage	+25°C		2	4		3	6	mV
	Full		3	6		4	8	mV
*Bias Current	+25°C		40	200		40	200	nA
	Full			400			400	nA
*Offset Current	+25°C		10	50		10	50	nA
	Full			100			100	nA
Input Resistance	+25°C	5	10		5	10		MΩ
Common Mode Range	Full	±10			±10			V
<b>TRANSFER CHARACTERISTICS</b>								
*Large Signal Voltage Gain (Note 1, 4)	Full	25K	50K		25K	50K		V/V
*Common Mode Rejection (Note 2)	Full	80	90		74	90		dB
Hold Mode Feedthrough Attenuation (Note 9)	Full		-76			-76		dB
Gain Bandwidth Product (Note 3)	+25°C		2.5			2.5		MHz
<b>OUTPUT CHARACTERISTICS</b>								
*Output Voltage Swing (Note 1)	Full	±10			±10			V
Output Current	+25°C	±15			±15			mA
Full Power Bandwidth (Note 3, 4)	+25°C		100			100		kHz
Output Resistance (D.C.)	+25°C		.15			.15		Ω
<b>TRANSIENT RESPONSE</b>								
Rise Time (Note 3, 5)	+25°C		50	75		50	75	ns
Overshoot (Note 3, 5)	+25°C		25	40		25	40	%
Slew Rate (Note 3, 6)	+25°C	5	7		3	7		V/μs
<b>DIGITAL INPUT CHARACTERISTICS</b>								
Digital Input Current (V <sub>IN</sub> = 0V)	Full			0.8			0.8	mA
Digital Input Current (V <sub>IN</sub> = +5.0V)	Full			20			20	μA
Digital Input Voltage (Low)	Full			0.8			0.8	V
Digital Input Voltage (High)	Full	2.0			2.0			V
<b>SAMPLE/HOLD CHARACTERISTICS</b>								
Acquisition Time to .1% 10V Step (Note 3)	+25°C		4			4		μs
Acquisition Time to .01% 10V Step (Note 3)	+25°C		5			5		μs
Aperture time (Note 10)	+25°C		30			30		ns
Effective Aperture Delay Time	+25°C		30			30		ns
Aperture Uncertainty	+25°C		5			5		ns
*Drift Current (Note 3, 8)	+25°C		5	50		5	50	pA
	Full		0.5	4.0		0.5	1.0	nA
*Charge Transfer (Note 8)	+25°C		5	10		5	10	pC
<b>POWER SUPPLY CHARACTERISTICS</b>								
*Supply Current (+)	+25°C		3.5	5.5		3.5	5.5	mA
*Supply Current (-)	+25°C		2.5	3.5		2.5	3.5	mA
*Power Supply Rejection	Full	-80	-90		-74	-90		dB

- NOTES:
- RL = 2kΩ
  - VCM = ±10VDC
  - AV = +1, RL = 2kΩ, CL = 50pF
  - VOUT = 20V peak-to-peak
  - VOUT = 200mV peak-to-peak
  - VOUT = 10.0V peak-to-peak
  - Derate Power Dissipation by 4.3mW/°C above +105°C Ambient Temperature
  - VIN = 0V
  - fIN ≤ 100kHz
  - Derived from computer simulation only; not tested.
- \*100% Tested for DASH 8

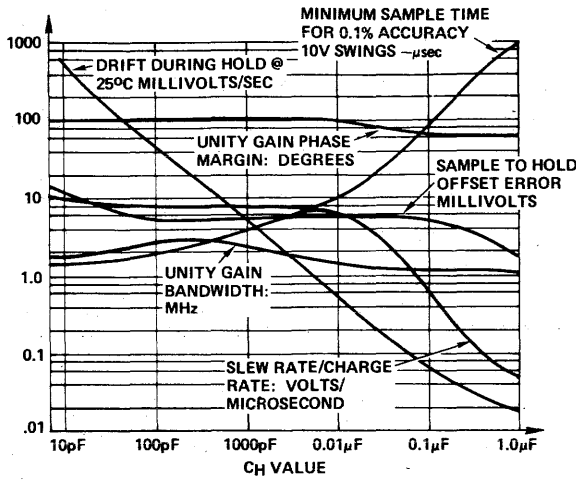


# PERFORMANCE CURVES

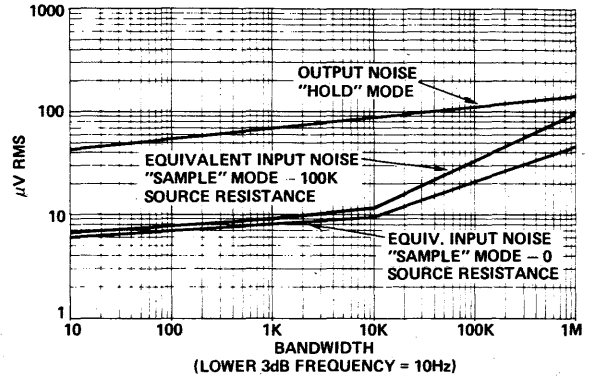


V<sub>SUPPLY</sub> = ±15VDC, T<sub>A</sub> = +25°C, C<sub>H</sub> = 1,000pF Unless Otherwise Specified

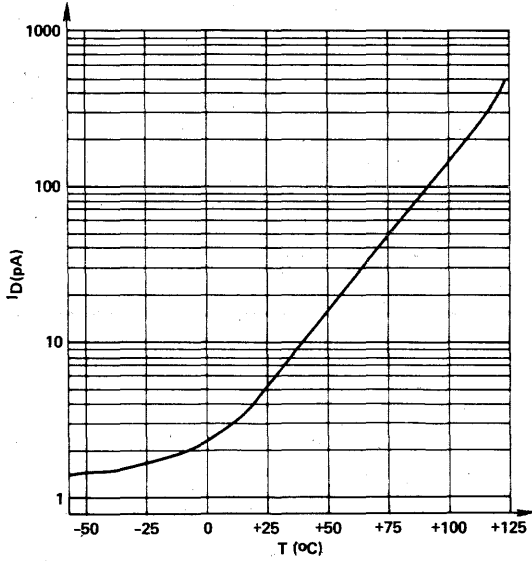
## TYPICAL SAMPLE AND HOLD PERFORMANCE AS A FUNCTION OF HOLDING CAPACITOR



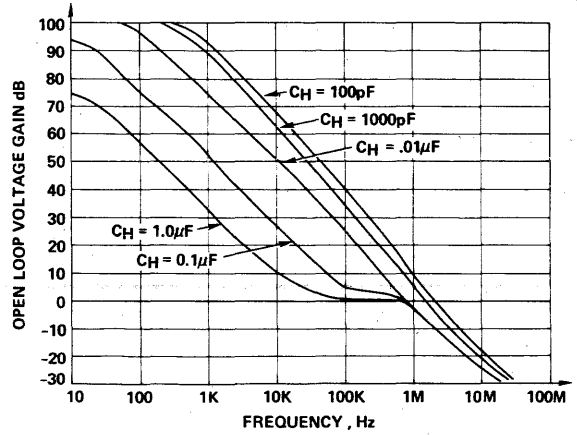
## BROADBAND NOISE CHARACTERISTICS



## DRIFT CURRENT VS. TEMPERATURE

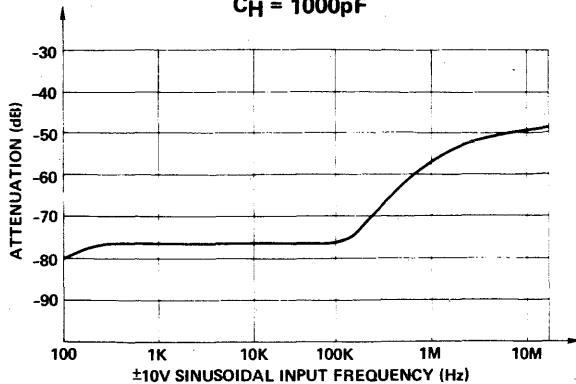


## OPEN LOOP FREQUENCY RESPONSE

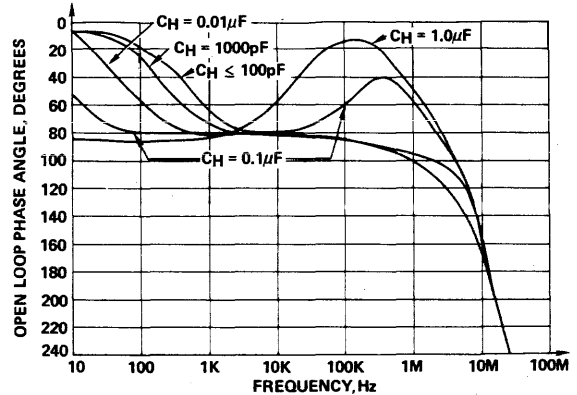


## HOLD MODE FEED THROUGH ATTENUATION

C<sub>H</sub> = 1000pF



## OPEN LOOP PHASE RESPONSE





# HA-5320

## High Speed Precision Monolithic Sample and Hold Amplifier

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• GAIN, dc <span style="float: right;">2 x 10<sup>6</sup> V/V</span></li> <li>• ACQUISITION TIME <span style="float: right;">1.0 μs (0.01%)</span></li> <li>• DROOP RATE <span style="float: right;">0.08 μV/μs (25°C)</span></li> <li style="padding-left: 100px;"><span style="float: right;">17 μV/μs (FULL TEMP)</span></li> <li>• APERTURE TIME <span style="float: right;">25ns</span></li> <li>• PEDESTAL ERROR <span style="float: right;">1.0 mV</span></li> <li>• INTERNAL HOLD CAPACITOR</li> <li>• FULLY DIFFERENTIAL INPUT</li> <li>• TTL COMPATIBLE</li> </ul>	<p>The HA-5320 was designed for use in precision, high speed data acquisition systems.</p> <p>The circuit consists of an input transconductance amplifier capable of providing large amounts of charging current, a low leakage analog switch, and an output integrating amplifier. The analog switch sees virtual ground as its load; therefore, charge injection on the hold capacitor is constant over the entire input/output voltage range. The pedestal voltage resulting from this charge injection can be adjusted to zero by use of the offset adjust inputs. The device includes a hold capacitor. However, if improved droop rate is required at the expense of acquisition time, additional hold capacitance may be added externally.</p> <p>This monolithic device is manufactured using the Harris dielectric isolation process, minimizing stray capacitance and eliminating SCR's. This allows higher speed and latch-free operation. The HA-5320 requires ±15V, and is available in a ceramic or plastic 14-pin DIP.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>• PRECISION DATA ACQUISITION SYSTEMS</li> <li>• D/A CONVERTER DEGLITCHING</li> <li>• AUTO-ZERO CIRCUITS</li> <li>• PEAK DETECTORS</li> </ul>	
PINOUT	FUNCTIONAL DIAGRAM
<p style="text-align: center;"><b>TOP VIEW</b></p>	

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# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage between V <sup>+</sup> and V <sup>-</sup> terminals	40V	Internal Power Dissipation	450mW (Note 2)
Differential Input Voltage	±24V	Operating Temperature Range	HA-5320-2/8 -55°C ≤ TA ≤ +125°C
Digital Input Voltage (Pin 14)	+8V, -15V	Storage Temperature Range	HA-5320-5 0°C ≤ TA ≤ +75°C
Output Current, continuous	±20mA (Note 1)		-65°C ≤ TA ≤ +150°C

## ELECTRICAL CHARACTERISTICS Test Conditions (unless otherwise specified)

V Supply = ±15V; C<sub>H</sub> - Internal; Digital Input (Pin 14), V<sub>AL</sub> = +0.8V (sample), V<sub>AH</sub> = +2.0V (hold).

PARAMETER	TEMP	HA-5320-2/8			HA-5320-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

### INPUT CHARACTERISTICS

Input Voltage Range	Full	±10			±10			V
Input Resistance	25°C	1	5		1	5		MΩ
Input Capacitance	25°C			3*			3	pF
Offset Voltage	25°C		0.2	0.5		0.5	1.0	mV
Bias Current	Full			2.0			1.5	mV
	25°C		70	200		100	300	nA
Offset Current	Full			200			300	nA
	25°C		30	100		30	300	nA
Common Mode Range	Full			100			300	nA
	25°C	±10			±10			V
CMRR (Note 3)	25°C	80	90		72	90		dB
Offset Voltage T.C.	Full		5	15		5	20	μV/°C

### TRANSFER CHARACTERISTICS

Gain, dC	25°C	10 <sup>6</sup>	2x10 <sup>6</sup>		3x10 <sup>5</sup>	2x10 <sup>6</sup>		V/V
Gain Accuracy, A <sub>v</sub> = +1	25°C		.5x10 <sup>-4</sup>			.5x10 <sup>-4</sup>		%FSR
Gain Accuracy, T.C.	Full		±.6			±.6		ppm/°C
Gain Bandwidth Product (Note 4, 5) CH = 100pF CH = 1000pF	25°C		2.0			2.0		MHz
			.18			.18		MHz

### OUTPUT CHARACTERISTICS

Output Voltage	Full	±10			±10			V
Output Current	25°C	±10			±10			mA
Full Power Bandwidth (Note 4, 6)	25°C		600			600		KHz
Output Resistance (Hold mode)	25°C		1.0			1.0		Ω
Total Output Noise, DC to 10 MHz	25°C		125	200		125	200	μV RMS
	25°C		125	200		125	200	μV RMS



PARAMETER	TEMP	HA-5320-2/8			HA-5320-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

TRANSIENT RESPONSE

Rise Time (Note 5)	25°C		100			100		nS
Overshoot (Note 5)	25°C		15			15		%
Slew Rate (Note 7)	25°C		45			45		V/μs

DIGITAL INPUT CHARACTERISTICS

Input Voltage (High), V <sub>AH</sub>	Full	2.0			2.0			V
Input Voltage (Low), V <sub>AL</sub>	Full			0.8			0.8	V
Input Current (V <sub>AL</sub> = 0V)	Full			4			4	μA
Input Current (V <sub>AH</sub> = +5V)	Full			0.1			0.1	μA

SAMPLE/HOLD CHARACTERISTICS

Acquisition Time (.1%) (Note 7)	25°C		.8	1.2		.8		μS
Acquisition Time (.01%) (Note 7)	25°C		1.0			1.0		μS
Aperture Time (Note 8)	25°C		25			25		nS
Effective Aperture Delay Time (See Glossary)	25°C		-25			-25		nS
Aperture Uncertainty	25°C		1			1		nS
Droop Rate	25°C		.08			.08		μV/us
Droop Rate	Full		17			1.2		μV/us
Drift Current (Note 9)	25°C		8			8		pA
Drift Current (Note 9)	Full		1.7			.12		nA
Charge Transfer (Note 9)	25°C		0.1			0.1		pC
Hold Mode Settling Time (.01%)	Full		185			185		nS
Hold Mode Feedthrough 10Vp-p, 100KHz	Full		2			2		mV

POWER SUPPLY CHARACTERISTICS

Positive Supply Voltage	Full	+14.5	+15	+16	+14.5	+15	+16	V
Negative Supply Voltage	Full	-14.5	-15	-16	-14.5	-15	-16	V
Positive Supply Current (Note 10)	25°C		11	13		11	13	mA
Negative Supply Current (Note 10)	25°C		-11	-13		-11	-13	mA
Power Supply Rejection V <sup>+</sup>	Full	80			80			dB
(Note 11) V <sup>-</sup>	Full	65			65			dB

NOTES:

- Internal Power Dissipation may limit Output Current below +20mA.
- Derate power dissipation by 4.3 mW/°C above 105°C ambient.
- V<sub>cm</sub> = +5V DC
- A<sub>v</sub> = +1; R<sub>L</sub> = 1KΩ; C<sub>L</sub> = 50pF
- V<sub>o</sub> = 200mVpp; R<sub>L</sub> = 2KΩ; C<sub>L</sub> = 50pF
- V<sub>o</sub> = 20Vpp.
- V<sub>o</sub> = 10V step; R<sub>L</sub> = 2KΩ; C<sub>L</sub> = 50pF
- Derived from computer simulation only; not tested.
- V<sub>IN</sub> = 0V, V<sub>AH</sub> = +3.5V, t<sub>r</sub> < 50ns (V<sub>AL</sub> to V<sub>AH</sub>).
- Specified for a zero differential input voltage between pins 1 and 2. Supply current will increase with differential input (as may occur in the Hold mode) to approximately ±28mA at 20V.
- Based on a one volt delta in each supply, ie: 15V ± 0.5 VDC.



# HI-5712/5712A

## High Performance 12 Bit Analog to Digital Converter

INTERFACE

Harris Semiconductor

### FEATURES

- MICROPROCESSOR COMPATIBLE
- CONVERSION TIME 10 $\mu$ sec MAX  
OVER TEMP.
- NO MISSING CODES OVER TEMPERATURE
- INTERNAL +10V REFERENCE
- INTERNAL CLOCK WITH EXTERNAL OVERRIDE CAPABILITY
- SERIAL OUTPUT
- TTL/CMOS COMPATIBLE
- TRISTATE PARALLEL OUTPUTS
- 40 PIN DIP
- MIL-STD-883 PROCESSING AVAILABLE

### APPLICATIONS

- MULTI-CHANNEL DATA ACQUISITION SYSTEMS
- STATUS MONITORING SYSTEMS
- PROCESS CONTROL SYSTEMS
- INSTRUMENTATION
- HIGH RELIABILITY DAS's

### DESCRIPTION

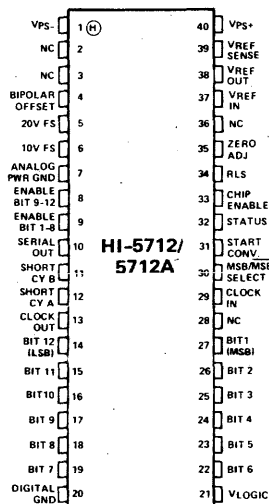
The HI-5712/5712A is a 12-bit successive approximation analog-to-digital converter (ADC) intended for high-speed, high-performance data conversion applications. An 8 $\mu$ s conversion time for an accurate 12 bit conversion with low gain and offset temperature coefficients are among its many features. Numerous functions can be software controlled to meet a variety of ADC requirements.

The highly flexible input design accepts user programmed unipolar and bipolar inputs of: 0 to +10V, 0 to +20V,  $\pm 5V$  and  $\pm 10V$  full scale signal levels. The internal precision +10V reference delivers up to 10mA of output current with ultra high temperature stability. This reference is intended for biasing the ADC reference input, although other configurations can be implemented. A remote sense line is provided for applications requiring usage of the precision reference elsewhere in the system.

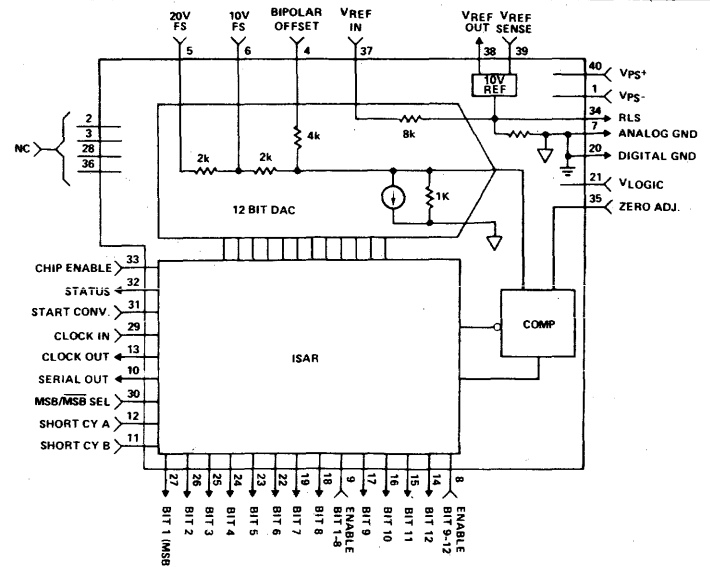
The output code select line and the short cycle control inputs are latched internally for microprocessor compatibility and provide selection of either binary or 2's complement output code, and resolution of 6, 8, 10 or 12 bits, respectively. A flexible interface is provided for 8, 12 and 16 bit systems via the chip select line and the word length control pins. The latter allows independent tri-state enabling of parallel output bits 1-8 and 9-12. A serial data output line is provided for applications requiring remote data transmission.

The HI-5712/5712A is manufactured with hermetically sealed leadless chip carriers (LCC's) mounted to both sides of a multi-layer ceramic substrate which results in a compact 40 pin dual-in-line package. The HI-5712A is intended for military, industrial and instrumentation applications. MIL-STD-883 class B and high reliability commercial grades are both available as standard products.

### PINOUT



### FUNCTIONAL DIAGRAM



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**ABSOLUTE MAXIMUM RATINGS (NOTE 1)**

Power Supply Inputs		Power Dissipation (Pd) 2 Watts	
V <sub>ps</sub> <sup>+</sup>	+20V	Operating Temperature Range	
V <sub>ps</sub> <sup>-</sup>	-20V	HI-5712-2, HI-5712A-2	55°C to +125°C
V <sub>LOGIC</sub>	+10V	HI-5712-5, HI-5712A-5	0°C to +75°C
V <sub>REF IN</sub> (Pin 37)	0V, V <sub>ps</sub> <sup>+</sup>	HI-5712-7, HI-5712A-7	0°C to +75°C (Hi Rel)
V <sub>REF SENSE</sub> (Pin 39)	0V, V <sub>ps</sub> <sup>+</sup>	HI-5712-8, HI-5712A-8	-55°C to +125°C (Hi Rel)
Digital Inputs	-1V, V <sub>LOGIC</sub>	Storage Temperature Range	65°C to +150°C

**ELECTRICAL CHARACTERISTICS**

(T<sub>A</sub> = +25°C, V<sub>ps</sub> = +15V, V<sub>ps</sub><sup>-</sup> = -15V, V<sub>LOGIC</sub> = +5V, V<sub>REF IN</sub> = Internal V<sub>REF</sub>, Full Scale = +10V, Conversion Speed = 9 μs TYP (Internal Clock), 12-BIT Conversion, Unless otherwise noted.)

PARAMETER	TEMP	HI-5712A-2 HI-5712-2		HI-5712A-5 HI-5712-5		UNITS			
		MIN	TYP	MAX	MIN		TYP	MAX	
RESOLUTION	Full			12		12	BITS		
NONLINEARITY	+25°C	HI-5712A		±1/4	±1/2		±1/4	±1/2	LSB
				Full	±1/4	±1/2		±1/4	±1/2
	+25°C	HI-5712		±1/4	±1/2		±1/4	±1/2	LSB
				Full	±1/2	±1		±1/2	±1
DIFFERENTIAL NONLINEARITY	+25°C	HI-5712A		±1/4	±1/2		±1/4	±1/2	LSB
				Full	±1/4	±1/2		±1/4	±1/2
	+25°C	HI-5712		±1/4	±1/2		±1/4	±1/2	LSB
				Full	±1/2	±1		±1/2	±1
NO MISSING CODES GUARANTEED OVER TEMPERATURE									
INHERENT QUANTIZATION ERROR	Full			±1/2			±1/2	LSB	
UNIPOLAR OFFSET ERROR (Note 2) (Adjustable to Zero)	+25°C		.3	.6		.3	.6	%FSR	
BIPOLAR OFFSET ERROR (Note 2) (Adjustable to Zero)	+25°C		.3	.6		.3	.6	%FSR	
GAIN ERROR (note 2) (Adjustable to Zero)	+25°C		.1	.3		.1	.3	%FSR	
ADJUSTMENT RANGE									
UNIPOLAR OFFSET	+25°C	±1	±2		±1	±2		%FSR	
BIPOLAR OFFSET	+25°C	±1	±2		±1	±2		%FSR	
GAIN	+25°C			.3			.3	%FSR	
TEMPERATURE STABILITY (With Internal V <sub>REF</sub> )									
UNIPOLAR OFFSET	Full		±2	±5		±2	±5	ppm FSR/°C	
DRIFT	Full		±4	±15		±4	±15	ppm FSR/°C	
BIPOLAR OFFSET	Full		±4	±10		±4	±10	ppm FSR/°C	
DRIFT	Full		±8	±25		±8	±25	ppm FSR/°C	
GAIN DRIFT	Full		±5	±10		±5	±10	ppm FSR/°C	
	Full		±10	±20		±10	±20	ppm FSR/°C	
NO MISSING CODES GUARANTEED OVER TEMPERATURE									



PARAMETER	TEMP	HI-5712A-2/-8 HI-5712-2/-8			HI-5712A-5/-7 HI-5712-5/-7			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
CONVERSION SPEED (Internal Clock)								
12 BIT	+25°C		9.0	10.0		9.0	10.0	μs
10 BIT	"		6.8	8.5		6.8	8.5	μs
8 BIT	"		5.6	7.0		5.6	7.0	μs
6 BIT	"		4.4	5.4		4.4	5.4	μs
MAXIMUM CONVERSION SPEED AT 12 BITS WITH EXTERNAL CLOCK (Note 3)	Full		6.5			6.5		μs

**ANALOG INPUT CHARACTERISTICS**

INPUT VOLTAGE RANGE UNIPOLAR	Full	10			10			V
	Full	20			20			V
BIPOLAR	Full	±5	±5		±5	±5		V
	Full	±10	±10		±10	±10		V
INPUT IMPEDANCE								
10V FS (PIN 6)	Full	1.6	2	2.4	1.6	2	2.4	KΩ
20V FS (PIN 5)	Full	3.2	4	4.8	3.2	4	4.8	KΩ
VREF IN (PIN 37)	Full	6.4	8	9.6	6.4	8	9.6	KΩ

**ANALOG OUTPUT CHARACTERISTICS**

VREF OUTPUT VOLTAGE	+25°C	9.970	10.000	10.030	9.970	10.000	10.030	V
VREF OUTPUT CURRENT	Full	10			10			mA
VREF OUTPUT TC	HI-5712A HI-5712	Full	±10	±15	±10	±15	±15	ppm FSR/°C

**DIGITAL INPUT CHARACTERISTICS**

INPUT VOLTAGE (Note 8)								
LOGIC 1	Full	3.3	2.7		3.3	2.7		V
LOGIC 0	Full		1.2	.8		1.2	.8	V
INPUT CURRENT (Note 8)								
LOGIC 1 (VCC)	Full	-25	0	+25	-25	0	+25	μA
LOGIC 0 (GND)	Full		-200	-400		-200	-400	μA
EXTERNAL CLOCK (Note 3)	Full			2.5			2.5	MHz

**DIGITAL OUTPUT CHARACTERISTICS**

OUTPUT VOLTAGE								
LOGIC 1 IOH = -800 A	Full	3.5	4.0		3.5	4.0		V
LOGIC 0 IOL = +3.2mA	Full		.2	.4		.2	.4	V
OUTPUT CURRENT								
LOGIC 1 VO = 3.5V	Full	-800	-1000		-800	-1000		μA
LOGIC 0 VO = .4V	Full	3.2	4.0		3.2	4.0		mA

**DIGITAL INPUT TIMING CHARACTERISTICS**

CHIP ENABLE TO START CONVERT tcd	Full	50			50			nsec
START CONVERT PULSE LOW tscL	Full	100			100			nsec
START CONVERT PULSE HIGH tscH	Full	50			50			nsec
CONTROL SETUP TIME ts	Full	100			100			nsec
CONTROL HOLD TIME th	Full	100			100			nsec
CLOCK INPUT LOW tpwL	Full	125			125			nsec
CLOCK INPUT HIGH tpwH	Full	150			150			nsec
CLOCK INPUT PERIOD tcl	Full	400			400			nsec
ENABLE 1-8, 9-12 PULSE WIDTH tem	Full	100			100			nsec



PARAMETER	TEMP	HI-5712A-2/-8 HI-5712-2/-8			HI-5712A-5/-7 HI-5712-5/-7			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	

DIGITAL OUTPUT TIMING CHARACTERISTICS (See Figure 6)

THREE STATE ENABLE DELAY	$t_{oe}$	Full		40	50		40	50	nsec
THREE STATE DISABLE DELAY	$t_{od}$	Full		60	100		60	100	nsec
START CONVERT TO STATUS DELAY	$t_{sd}$	Full		70	100		70	100	nsec
START CONVERT TO CLOCK OUT DELAY	$t_{scd}$			200	500		200	500	nsec
CLOCK TO SERIAL OUT DELAY	$t_{psd}$	Full	100	150	200	100	150	200	nsec
LAST CLOCK TO STATUS DELAY	$t_{scdt}$	Full	50	75	100	50	75	100	nsec
PARALLEL DATA TO STATUS DELAY	$t_{ds}$	Full	50	75		50	75		nsec
LAST SERIAL BIT TO STATUS DELAY	$t_{da}$	Full	50	75		50	75		nsec
CLOCK INPUT TO CLOCK OUT DELAY	$t_{dcl}$	Full		25	50		25	50	nsec

PARALLEL DATA OUTPUT CODES

UNIPOLAR (Note 4)	Positive True Binary
BIPOLAR (Note 4)	Positive True Offset Binary
	Positive True Two's Complement Binary
SERIAL DATA OUTPUT CODE	Positive True NRZ Code

POWER SUPPLY REQUIREMENTS (Note 5)

$V_{ps+}$	Full	+13.5	+15	+16.5	+13.5	+15	+16.5	V
$V_{ps-}$	Full	-13.5	-15	-16.5	-13.5	-15	-16.5	V
$V_{LOGIC}$	Full	+4.5	+5	+5.5	+4.75	+5	+5.25	V
$I_{ps+}$	Full		27	35		27	35	mA
$I_{ps-}$	Full		42	50		42	50	mA
$I_{LOGIC}$	Full		4.5	15		4.5	15	mA

POWER SUPPLY SENSITIVITY (Note 6)

$V_{ps+} = +13.5V$ to $+16.5V$ $V_{ps-} = -15V$ , $V_{LOGIC} = +5V$								ppm of FSR/ % $\Delta$ P.S.
UNIPOLAR OFFSET			2	5		2	5	
BIPOLAR OFFSET			2	4		2	4	
GAIN			1	3		1	3	
$V_{ps-} = -13.5V$ to $-16.5V$ $V_{ps+} = +15V$ , $V_{LOGIC} = +5V$								ppm of FSR/ % $\Delta$ P.S.
UNIPOLAR OFFSET			2	5		2	5	
BIPOLAR OFFSET			2	4		2	4	
GAIN			1	3		1	3	
$V_{LOGIC} = +4.5V$ to $+5.5V$ $V_{ps+} = +15V$ , $V_{ps-} = -15V$ CONVERSION SPEED (12 Bit with Internal Clock)								%
			$\pm 5$	$\pm 10$		$\pm 5$	$\pm 10$	





- NOTES: 1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. See Figure 2 for connections. The initial errors are adjustable to zero by using external trim potentiometers as shown in Figure 3, and 4.
3. The HI-5712A will operate at these speeds (for 12 bit conversion), but parametric performance is not guaranteed.
4. See operating instructions for details.
5. After 60 seconds warm-up.
6. See definitions.
7. These terminals will be used in the future for additional functions. Do not make connections to these pins in your system.
8. TTL compatibility guaranteed.

### PIN FUNCTIONS AND DESCRIPTIONS

PIN	SYMBOL	DESCRIPTION															
1	V <sub>ps-</sub>	-15V Power Supply Terminal															
2	NC	No Connection See Note 7															
3	NC	No Connection See Note 7															
4	BIPOLAR OFFSET	Connect to VREF for Bipolar Input Mode. See Operating Instructions for Details.															
5	20V FS	20V Full Scale Analog Input															
6	10V FS	10V Full Scale Analog Input															
7	ANALOG GND	Analog Power Supply Return															
8	ENABLE BIT 9-12	Output "Three State" Control. An Input "0" Enables Bits 9 through 12, whereas a "1" Switches these Bits to a High Impedance State.															
9	ENABLE BIT 1-8	Output "Three State" Control. An Input "0" Enables Bits 1 through 8, whereas a "1" Switches these Bits to a High Impedance State.															
10	SERIAL OUT	NRZ Serial Data Output. To be used in Conjunction with Clock Out for Remote Data Transmission															
11	SHORT CY B	See Description for Pin 12															
12	SHORT CY A	Digital Inputs Applied to short cycle A and B selects a conversion of 6, 8, 10, or 12-bits:															
		<table border="1"> <thead> <tr> <th>BITS</th> <th>SHORT CY A</th> <th>SHORT CY B</th> </tr> </thead> <tbody> <tr> <td>6</td> <td>0</td> <td>0</td> </tr> <tr> <td>8</td> <td>0</td> <td>1</td> </tr> <tr> <td>10</td> <td>1</td> <td>0</td> </tr> <tr> <td>12</td> <td>1</td> <td>1</td> </tr> </tbody> </table>	BITS	SHORT CY A	SHORT CY B	6	0	0	8	0	1	10	1	0	12	1	1
BITS	SHORT CY A	SHORT CY B															
6	0	0															
8	0	1															
10	1	0															
12	1	1															

PIN	SYMBOL	DESCRIPTION
13	CLOCK OUT	SAR Clock Output. Used for Decoding Serial Out Data
14	BIT 12	Output Data Bit (LSB)
15	BIT 11	Output Data Bit
16	BIT 10	Output Data Bit
17	BIT 9	Output Data Bit
18	BIT 8	Output Data Bit
19	BIT 7	Output Data Bit
20	DIGITAL GND	Digital Power Supply Return
21	V <sub>LOGIC</sub>	+5V Power Supply Terminal
22	BIT 6	Output Data Bit
23	BIT 5	Output Data Bit
24	BIT 4	Output Data Bit
25	BIT 3	Output Data Bit
26	BIT 2	Output Data Bit
27	BIT 1	Output Data Bit (MSB)
28	NC	No Connection. See Note 7.
29	CLOCK IN	An External Clock Signal Applied to this Input Overrides the Internal Clock.
30	MSB/ $\overline{\text{MSB}}$ SEL	Digital Input Pin. A "1" Applied to this Terminal Selects a Straight Binary or Offset Binary Output Code. A "0" Inverts the MSB to Yield a 2's Complement Binary Output Code.
31	START CONV	Digital Input Pin. A High to Low Transition Initiates the ADC Conversion Cycle.
32	STATUS	Digital Output Pin. A "1" Indicates that the ADC is Busy, While a "0" Denotes that Conversion is Completed and Data is Ready for Retrieval.



PIN	SYMBOL	DESCRIPTION
33	CHIP ENABLE	Digital Input Pin. A "1" Forces the Output Data, Serial Out and Status Terminals to a High Impedance State and the ADC is Disabled. A "0" Enables these ADC Functions.
34	RLS	Reference Low Sense.
35	ZERO ADJ	External Zero Adjustment Pin, See Operating Instructions for Details.

PIN	SYMBOL	DESCRIPTION
36	NC	No Connection. See Note 7.
37	VREF IN	+10V Reference Input to ADC.
38	VREF OUT	Internal +10V Reference Output, Normally Connected to VREF IN (Pin 37).
39	VREF SENSE	Internal +10V Reference Sensing Terminal, Normally connected to VREF Out (Pin 38). See Operating Instructions for Details.
40	V <sub>ps+</sub>	+15V Power Supply Terminal.

**APPLYING THE HI-5712/5712A**

**OPERATING INSTRUCTIONS**

Conventional ADC systems provide maximum performance when the analog and digital ground lines are tied together at the ADC terminals. This minimizes analog interference due to digital switching noise. For optimum performance, this external grounding procedure should be followed in HI-5712/5712A installations to reinforce the unit's internal analog-to-digital ground connections. Under no circumstances should the Reference Low Sense (RLS) terminal (Pin 34) be connected to system ground.

In practice, the Reference Low Sense (RLS) terminal (Pin 34) normally is connected to zero adjust (or error amplifier) input terminal (Pin 35), either directly or through an appropriate resistor network. See figures 3 and 4.

On the HI-5712/5712A substrate, the power supply lines to each active component are bypassed to ground with 0.01 μF chip capacitors for high frequency noise rejection.

For best accuracy, the grounding and decoupling schemes shown in Figures 3 and 4 are recommended. The 10 μF bypass capacitors shown should be connected as close as possible to the HI-5712/5712A, preferably at the device pin.

For applications where usage of potentiometers is highly undesirable, the trim pots shown in Figures 3 and 4 can either be deleted or replaced by precision fixed resistors. (Delete R<sub>3</sub> and R<sub>4</sub>; replace R<sub>1</sub> with 25 ohms). When precision fixed resistors are used, the initial offset error and gain error contributions are as specified in page 2.

**NOTE:** The HI-5712/5712A may latch up if the device is enabled before applying power. Disabling the device following power turn on will remedy this situation. Care supplies do not excessively overshoot their final value during turn on.

**CONTROL AND INTERFACE**

The HI-5712/5712A features a versatile set of controlling functions which allows a wide variety of applications, including microprocessor bus interfacing.

When the chip enable is set to low, the internal registers are enabled, and the output data lines can be enabled via the output enable control lines. The conversion cycle is initiated at the falling edge of the start conversion pulse. At this time, the MSB/MSB Select, Short Cycle A, and Short Cycle B control information is latched into the internal registers. The status line is also forced into an active high state indicating that a conversion is taking place. At the

end of the conversion cycle the status line will be set to low to signify that the data is ready at the tri-state buffers. The various timing relationships are shown in Figure 1.

There are two distinct modes of operation, namely, continuous conversion and single step conversion. Continuous conversion can be easily achieved by connecting the Status line to the Start Convert pin. In this application, an indcision state may occur during the initial power-on conditions. Normal operation is restored by pulsing the chip enable pin to logic high for a period greater than 100 ns.

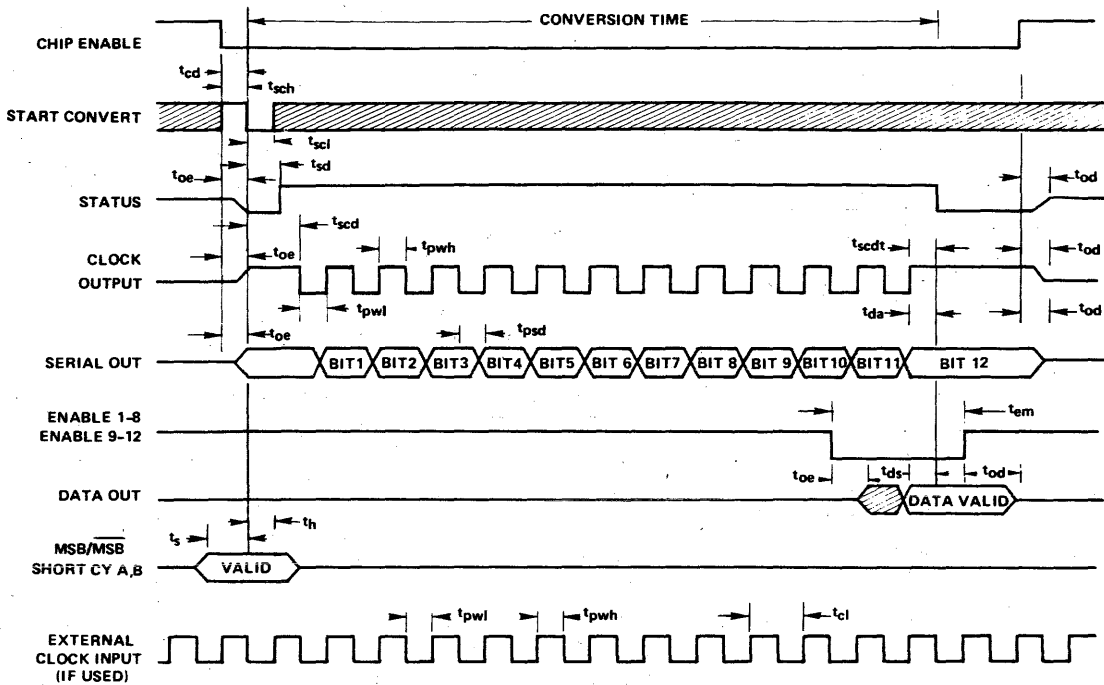


FIGURE 1. HI-5712/5712A TIMING DIAGRAM

REMOTE DATA TRANSMISSION

The Serial Data Out is mainly used for remote data transmission, where only a limited number of wires are available.

Serial Output is bit by bit (MSB first, LSB last) in a NRZ (nonreturn-to-zero) format. It changes state only at the positive going edges of the Clock Out, and remains valid during the whole clock period. Parallel data can be constructed by clocking the serial data into a receiving shift register.

In order to minimize transmission error, the negative-going edge of the clock should be used to clock data into the remote shift register. The parallel data will be valid once the status line returns to low. The clocking scheme is shown in Figure 1.

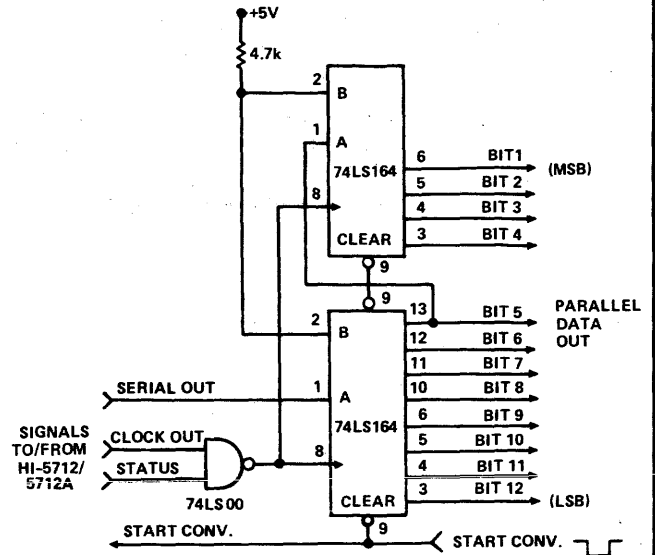


FIGURE 2. DECODING SERIAL DATA OUT



HI-5712/5712A CALIBRATION CHART

OPERATING MODE	ANALOG INPUT CONNECTION	R2 BIAS RESISTOR	MSB/MSB SELECT PIN 30	OFFSET ADJUST ANALOG INPUT VOLTAGE	ADJUST R3 FOR DITHER BETWEEN CODES	GAIN ADJUST ANALOG INPUT VOLTAGE	ADJUST R1 FOR DITHER BETWEEN CODES	LSB WEIGHT
UNIPOLAR STRAIGHT BINARY 0V to +10V	10VFS PIN 6	667Ω	HIGH	+1.22mV	0000 0000 0000 0000 0000 0001	+9.9963V	1111 1111 1110 1111 1111 1111	2.44mV
UNIPOLAR STRAIGHT BINARY 0V to +20V	20VFS PIN 5	800Ω	HIGH	+2.44mV	0000 0000 0000 0000 0000 0001	+19.9927V	1111 1111 1110 1111 1111 1111	4.88mV
BIPOLAR OFFSET BINARY -5V to +5V	10VFS PIN 6	580Ω	HIGH	-4.9988V	0000 0000 0000 0000 0000 0001	+4.9963V	1111 1111 1110 0111 1111 1111	2.44mV
BIPOLAR OFFSET BINARY -10V to +10V	20V FS PIN 5	667Ω	HIGH	-9.9976V	0000 0000 0000 0000 0000 0001	+9.9927V	1111 1111 1110 1111 1111 1111	4.88mV
BIPOLAR 2's COMPLEMENT -5V to +5V	10V FS PIN 5	580Ω	LOW	-4.9988V	1000 0000 0000 1000 0000 0001	+4.9963V	0111 1111 1110 0111 1111 1111	2.44mV
BIPOLAR 2's COMPLEMENT 10V to +10V	20V FS PIN 6	667Ω	LOW	-9.9976V	1000 0000 0000 1000 0000 0001	+9.9927V	0111 1111 1110 0111 1111 1111	4.88mV

CALIBRATION PROCEDURE- Refer to Calibration Chart and to Figures 3 and 4 for appropriate analog input connections, value of bias resistor, and MSB/MSB select.

STEP 1 OFFSET ADJUSTMENT

- Set analog input to the appropriate value for offset adjustment.
- Adjust R3 for dither between codes shown in calibration chart.

STEP 2 GAIN ADJUSTMENT

- Set analog input to the appropriate value for gain adjustment.
- Adjust R1 for dither between codes shown in calibration chart.

NOTE: This calibration procedure insures that the transfer characteristic produced by connecting the midpoints of all quantization intervals passes through the origin.

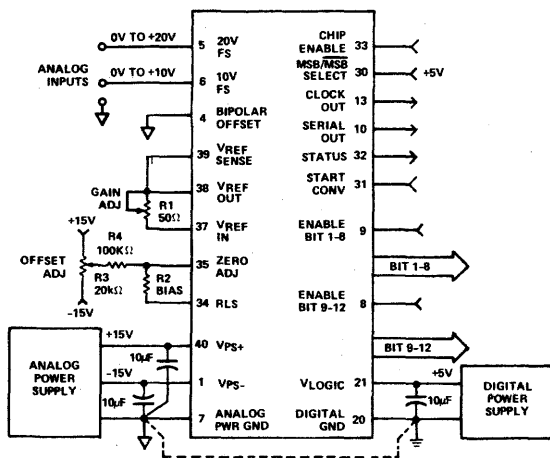


FIGURE 3. UNIPOLAR INPUT CONNECTIONS - STRAIGHT BINARY OUTPUT CODE

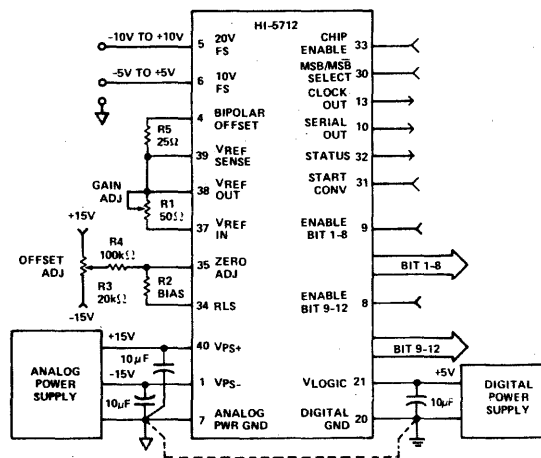


FIGURE 4. BIPOLAR INPUT CONNECTIONS

## Fast, Complete 12-Bit A/D Converter with Microprocessor Interface

### FEATURES

- COMPLETE 12 BIT A/D CONVERTER WITH REFERENCE AND CLOCK.
- FULL 8-, 12- or 16- BIT MICROPROCESSOR BUS INTERFACE.
- 150 nS BUS ACCESS TIME
- NO MISSING CODES OVER TEMPERATURE
- MINIMAL SETUP TIME FOR CONTROL SIGNALS
- 25  $\mu$ S MAXIMUM CONVERSION TIME
- LOW NOISE, VIA CURRENT-MODE SIGNAL TRANSMISSION BETWEEN CHIPS
- BYTE ENABLE/SHORT CYCLE ( $A_0$  INPUT)
  - GUARANTEES BREAK - BEFORE - MAKE ACTION, ELIMINATING BUS CONTENTION DURING READ OPERATION. LATCHED BY THE START CONVERT INPUT (TO SET THE CONVERSION LENGTH).
- IMPROVED SECOND SOURCE FOR AD574A AND HS574
- $\pm 12V$  TO  $\pm 15V$  OPERATION

### APPLICATIONS

- MILITARY AND INDUSTRIAL DATA ACQUISITION SYSTEMS
- ELECTRONIC TEST AND SCIENTIFIC INSTRUMENTATION.
- PROCESS CONTROL SYSTEMS.

### DESCRIPTION

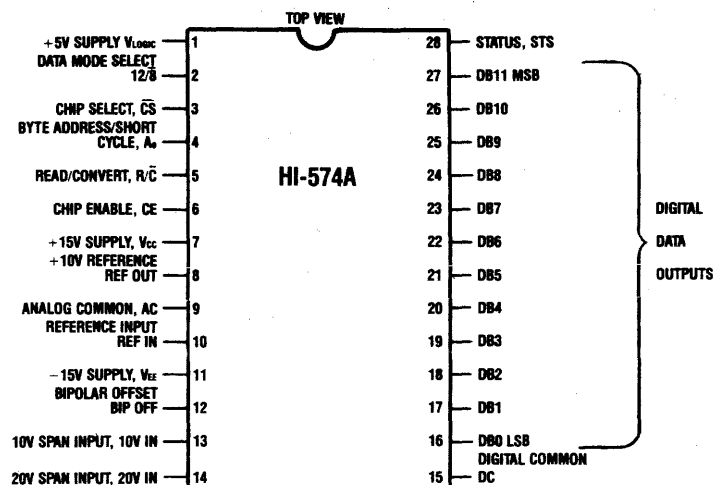
The HI-574A is a complete 12 bit Analog-to-Digital Converter, including a +10V reference, clock, three-state outputs and a digital interface for microprocessor control. Successive approximation conversion is performed by two monolithic dice housed in a 28-pin package. The bipolar analog die features the Harris Dielectric Isolation process, which provides enhanced AC performance and freedom from latch-up.

Custom design of each IC (bipolar analog and CMOS digital) has yielded improved performance over existing versions of this converter. The voltage comparator features high PSRR plus a high speed current-mode latch, and provides precise decisions down to 0.1 LSB of input overdrive. More than 2X reduction in noise has been achieved by using current instead of voltage for transmission of all signals between the analog and digital IC's. Also, the clock oscillator is current-controlled for excellent stability over temperature. The oscillator is trimmed for a nominal conversion time of  $20 \pm 1 \mu$ s.

The HI-574A offers standard unipolar and bipolar input ranges, laser trimmed for specified linearity, gain and offset accuracy. The buried zener reference circuit is trimmed for minimum temperature coefficient.

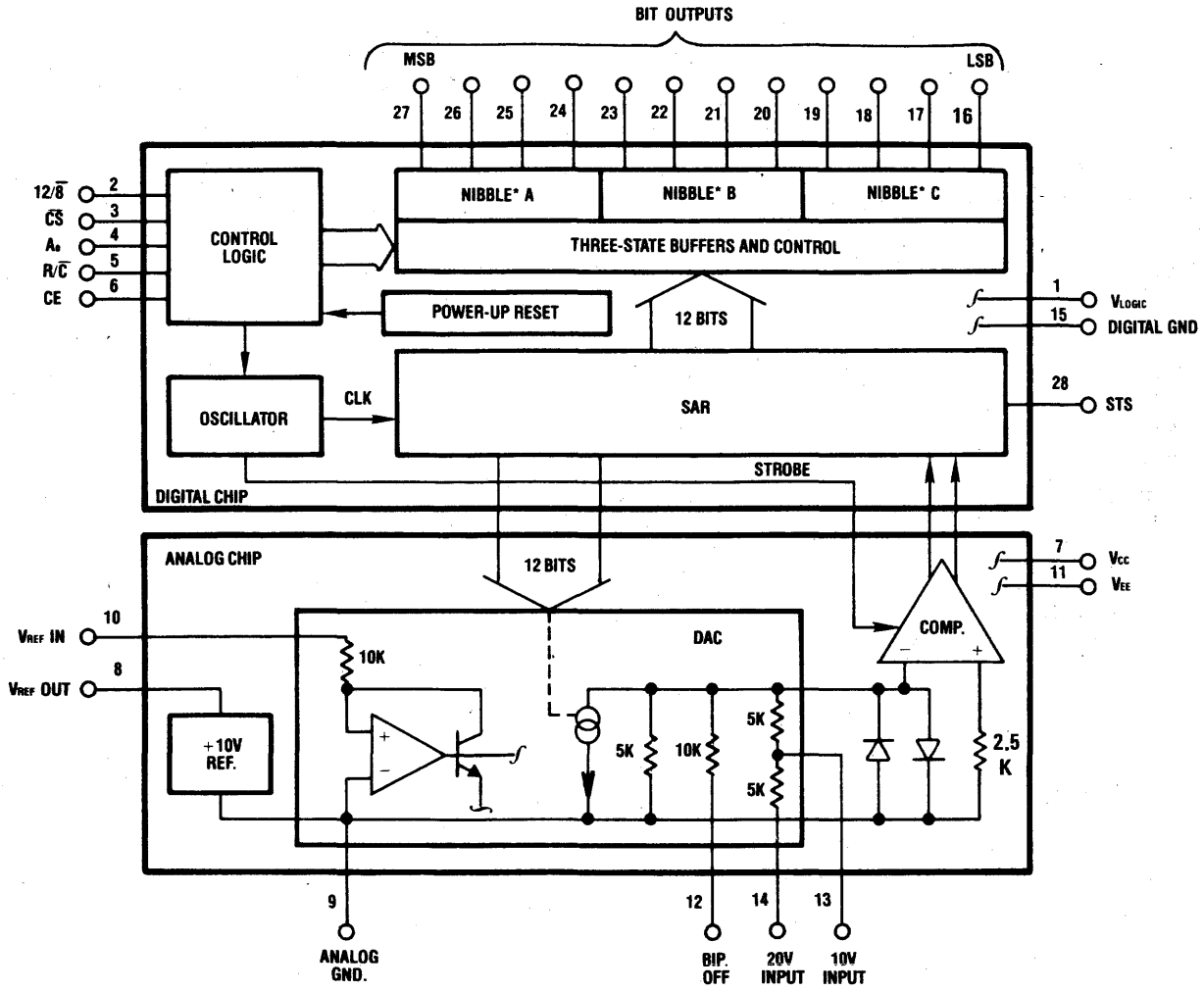
Power requirements are +5V and  $\pm 12V$  to  $\pm 15V$ , with typical dissipation of 515 mW. Three electrical grades each are offered for the commercial and military temperature ranges. All models are packaged in a 28 pin side-brazed, ceramic DIP. In addition, parts are available with screening per MIL-STD-883 Method 5004, Class B.

### PINOUT



CAUTION: These devices are sensitive to electronic discharge. Proper I.C. handling procedures should be followed.

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("NIBBLE" IS A 4 BIT DIGITAL WORD.)

HI-574A BLOCK DIAGRAM

# SPECIFICATIONS



(Typical @ +25°C with  $V_{CC} = +15V$  or  $+12V$ ,  $V_{LOGIC} = +5V$ ,  $V_{EE} = -15V$  or  $-12V$  unless otherwise specified)

## DC AND TRANSFER ACCURACY SPECIFICATIONS

MODEL	HI-574AJ	HI-574AK	HI-574AL	UNITS
Temperature Range	0 TO +75			°C
Resolution (max)	12	12	12	Bits
Linearity Error 25°C (max)	±1	±1/2	±1/2	LSB
0°C to +75°C (max)	±1	±1/2	±1/2	LSB
Differential Linearity Error (Maximum resolution for which no missing codes is guaranteed) 25°C	11	12	12	Bits
$T_{min}$ to $T_{max}$	11	12	12	Bits
Unipolar Offset (max) (Adjustable to zero)	±2	±2	±2	LSB
Bipolar Offset (max) (Adjustable to zero)	±10	±4	±4	LSB
Full Scale Calibration Error 25°C (max), with fixed 50 Ω resistor from REF OUT to REF IN (Adjustable to zero)	0.3	0.3	0.3	% of F.S.
$T_{min}$ to $T_{max}$ (No adjustment at +25°C)	0.5	0.4	0.35	% of F.S.
(With adjustment to zero at +25°C)	0.22	0.12	0.05	% of F.S.
Temperature Coefficients Guaranteed max change, $T_{min}$ to $T_{max}$ (Using internal reference)				
Unipolar Offset	±2 (10)	±1 (5)	±1 (5)	LSB (ppm/°C)
Bipolar Offset	±2 (10)	±1 (5)	±1 (5)	LSB (ppm/°C)
Full Scale Calibration	±9 (45)	±5 (25)	±2 (10)	LSB (ppm/°C)
Power Supply Rejection Max change in Full Scale Calibration $+13.5V < V_{CC} < +16.5V$ or $+11.4V < V_{CC} < +12.6V$ $+4.5V < V_{LOGIC} < +5.5V$ $-16.5V < V_{EE} < -13.5V$ or $-12.6V < V_{EE} < -11.4V$	±2 ±1/2 ±2	±1 ±1/2 ±1	±1 ±1/2 ±1	LSB LSB LSB
Analog Inputs Input Ranges Bipolar	-5 to +5 -10 to +10			Volts Volts
Unipolar	0 to +10 0 to +20			Volts Volts
Input Impedance 10 Volt Span 20 Volt Span	5K, ± 25% 10K, ± 25%			Ohms Ohms
Power Supplies Operating Voltage Range $V_{LOGIC}$ $V_{CC}$ $V_{EE}$	+4.5 to +5.5 +11.4 to +16.5 -11.4 to -16.5			Volts Volts Volts
Operating Current $I_{LOGIC}$ $I_{CC}$ $I_{EE}$	7 TYP, 15 MAX 11 TYP, 15 MAX 21 TYP, 28 MAX			mA mA mA
Power Dissipation	515 TYP, 720 MAX			mW
Internal Reference Voltage Output current, <sup>1</sup> available for external loads (External load should not change during conversion).	+10.00 ± 0.1 MAX 2.0 MAX			Volts mA

<sup>1</sup> When supplying an external load and operating on ±12V supplies, a buffer amplifier must be provided for the Reference Output.

**SPECIFICATIONS**



(Typical @ +25°C with V<sub>CC</sub> = +15V or +12V, V<sub>LOGIC</sub> = +5V, V<sub>EE</sub> = -15V or -12V unless otherwise specified)

**DC AND TRANSFER ACCURACY SPECIFICATIONS**

MODEL (Applicable to /883B versions also)	HI-574AS	HI-574AT	HI-574AU	UNITS
Temperature Range	-55 TO +125			°C
Resolution (max)	12	12	12	Bits
Linearity Error 25°C (max) -55°C to +125°C (max)	±1 ±1	±1/2 ±1	±1/2 ±1	LSB LSB
Differential Linearity Error (Maximum resolution for which no missing codes is guaranteed) 25°C T <sub>min</sub> to T <sub>max</sub>	11 11	12 12	12 12	Bits Bits
Unipolar Offset (max) (Adjustable to zero)	±2	±2	±2	LSB
Bipolar Offset (max) (Adjustable to zero)	±10	±4	±4	LSB
Full Scale Calibration Error 25°C (max), with fixed 50 Ω resistor from REF OUT to REF IN (Adjustable to zero) T <sub>min</sub> to T <sub>max</sub> (No adjustment at +25°C) (With adjustment to zero at +25°C)	0.3 0.8 0.5	0.3 0.6 0.25	0.3 0.4 0.12	% of F.S. % of F.S. % of F.S.
Temperature Coefficients Guaranteed max change, T <sub>min</sub> to T <sub>max</sub> (Using internal reference) Unipolar Offset  Bipolar Offset  Full Scale Calibration	±2 (5) ±4 (10) ±20 (50)	±1 (2.5) ±2 (5) ±10 (25)	±1 (2.5) ±1 (2.5) ±5 (12.5)	LSB (ppm/°C) LSB (ppm/°C) LSB (ppm/°C)
Power Supply Rejection Max change in Full Scale Calibration +13.5V < V <sub>CC</sub> < +16.5V or +11.4V < V <sub>CC</sub> < +12.6V +4.5V < V <sub>LOGIC</sub> < +5.5V -16.5V < V <sub>EE</sub> < -13.5V or -12.6V < V <sub>EE</sub> < -11.4V	±2 ±1/2 ±2	±1 ±1/2 ±1	±1 ±1/2 ±1	LSB LSB LSB
Analog Inputs Input Ranges Bipolar  Unipolar  Input Impedance 10 Volt Span 20 Volt Span	-5 to +5 -10 to +10  0 to +10 0 to +20  5K Ω, ± 25% 10K Ω, ± 25%			Volts Volts  Volts Volts  Ohms Ohms
Power Supplies Operating Voltage Range V <sub>LOGIC</sub> V <sub>CC</sub> V <sub>EE</sub>  Operating Current I <sub>LOGIC</sub> I <sub>CC</sub> I <sub>EE</sub>	+4.5 to +5.5 +11.4 to +16.5 -11.4 to -16.5  7 TYP, 15 MAX 11 TYP, 15 MAX 21 TYP, 28 MAX			Volts Volts Volts  mA mA mA
Power Dissipation	515 TYP, 720 MAX			mW
Internal Reference Voltage Output current, <sup>1</sup> available for external loads (External load should not change during conversion).	+10.00 ± 0.1 (MAX) 2.0 MAX			Volts mA

<sup>1</sup> When supplying an external load and operating on ±12V supplies, a buffer amplifier must be provided for the Reference Output.



**DIGITAL CHARACTERISTICS<sup>1</sup>**  
(ALL MODELS, OVER FULL TEMP. RANGE)

	MIN	TYP	MAX
Logic Inputs (CE, $\overline{CS}$ , R/ $\overline{C}$ , AO, 12/ $\overline{8}$ )			
Logic "1"	+2.4V <sup>2</sup>		+5.5V
Logic "0"	-0.5V		+0.8V
Current	-5 $\mu$ A	0.1 $\mu$ A	+5 $\mu$ A
Capacitance		5pF	
Logic Outputs (DB11-DB0, STS)			
Logic "0" (I <sub>SINK</sub> — 1.6mA)			+0.4V
Logic "1" (I <sub>SOURCE</sub> — 500 $\mu$ A)	+2.4V		
Leakage (High - Z State, DB11-DB0 ONLY)	-5 $\mu$ A	0.1 $\mu$ A	+5 $\mu$ A
Capacitance		5pF	

<sup>1</sup> See "HI-574A Timing Specifications" for a detailed listing of digital timing parameters.<sup>2</sup> Although this guaranteed threshold is higher than standard TTL (+2.0V), bus loading is much less, i.e., typical input current is only 0.25% of a TTL load.**ABSOLUTE MAXIMUM RATINGS**

(Specifications apply to all grades, except where noted)

V <sub>CC</sub> to Digital Common	0 to +16.5V
V <sub>EE</sub> to Digital Common	0 to -16.5V
V <sub>LOGIC</sub> to Digital Common	0 to +7V
Analog Common to Digital Common	±1V
Control Inputs (CE, $\overline{CS}$ , A <sub>0</sub> , 12/ $\overline{8}$ , R/ $\overline{C}$ ) to Digital Common	-0.5V to V <sub>LOGIC</sub> +0.5V
Analog Inputs (REF IN, BIP OFF, 10V <sub>IN</sub> ) to Analog Common	±16.5V

20V <sub>IN</sub> to Analog Common	±24V
REF OUT	Indefinite short to common Momentary short to V <sub>CC</sub>
Chip Temperature (J, K, L grades)	100°C
(S, T, U grades)	150°C
Power Dissipation	1000mW
Lead Temperature, Soldering	300°C, 10 sec.
Storage Temperature	-65°C to +150°C
Thermal Resistance, $\Theta_{JA}$	60°C/W

**HI-574A ORDERING GUIDE**

MODEL	TEMP. RANGE	LINEARITY ERROR MAX (T <sub>MIN</sub> to T <sub>MAX</sub> )	RESOLUTION (NO MISSING CODES, T <sub>MIN</sub> to T <sub>MAX</sub> )	FULL SCALE TC (PPM/°C MAX)
HI1-574AJD	0 to 75°C	±1 LSB	11 Bits	50.0
HI1-574AKD	0 to 75°C	±1/2 LSB	12 Bits	27.0
HI1-574ALD	0 to 75°C	±1/2 LSB	12 Bits	10.0
HI1-574ASD	-55 to +125°C	±1 LSB	11 Bits	50.0
HI1-574ASD/883B	-55 to +125°C	±1 LSB	11 Bits	50.0
HI1-574ATD	-55 to +125°C	±1 LSB	12 Bits	25.0
HI1-574ATD/883B	-55 to +125°C	±1 LSB	12 Bits	25.0
HI1-574AUD	-55 to +125°C	±1 LSB	12 Bits	12.5
HI1-574AUD/883B	-55 to +125°C	±1 LSB	12 Bits	12.5

**DEFINITIONS OF SPECIFICATIONS****LINEARITY ERROR**

Linearity error refers to the deviation of each individual code from a line drawn from "zero" through "full scale". The point used as "zero" occurs 1/2LSB (1.22mV for 10 volt span) before the first code transition (all zeros to only the LSB "on"). "Full scale" is defined as a level 1 1/2LSB beyond the last code transition (to all ones). The deviation of a code from the true straight line is measured from the middle of each particular code.

The HI-574AK, AL, AT, and AU grades are guaranteed for maximum nonlinearity of ±1/2LSB. For these grades, this means that an analog value which falls exactly in the center of a given code width will result in the correct digital output code. Values nearer the upper or lower

transition of the code width may produce the next upper or lower digital output code. The HI-574AJ and AS grades are guaranteed to ±1LSB max error. For these grades, an analog value which falls within a given code width will result in either the correct code for that region or either adjacent one.

Note that the linearity error is not user-adjustable.

**DIFFERENTIAL LINEARITY ERROR  
(NO MISSING CODES)**

A specification which guarantees no missing codes requires that every code combination appear in a monotonic increasing sequence as the analog input level is increased. Thus every code must have a finite width. For the HI-574AK, AL, AT, and AU grades, which



guarantee no missing codes to 12-bit resolution, all 4096 codes must be present over the entire operating temperature ranges. The HI-574AJ and AS grades guarantee no missing codes to 11-bit resolution over temperature; this means that all code combinations of the upper 11 bits must be present; in practice very few of the 12-bit codes are missing.

#### UNIPOLAR OFFSET

The first transition should occur at a level  $\frac{1}{2}$ LSB above analog common. Unipolar offset is defined as the deviation of the actual transition from that point. This offset can be adjusted as discussed on the following pages. The unipolar offset temperature coefficient specifies the maximum change of the transition point over temperature, with or without external adjustment.

#### BIPOLAR OFFSET

Similarly, in the bipolar mode, the major carry transition (0111 1111 1111 to 1000 0000 0000) should occur for an analog value  $\frac{1}{2}$ LSB below analog common. The bipolar offset error and temperature coefficient specify the initial deviation and maximum change in the error over temperature.

#### FULL SCALE CALIBRATION ERROR

The last transition (from 1111 1111 1110 to 1111 1111 1111) should occur for an analog value  $\frac{1}{2}$ LSB below the nominal full scale (9.9963 volts for 10.000 volts full scale). The full scale calibration error is the deviation of the actual level at the last transition from the ideal level. This error, which is typically 0.05 to 0.1% of full scale, can be trimmed out as shown in Figure 5. The full scale calibration error over temperature is given with and without the initial error trimmed out. The temperature coefficients for each grade indicate the maximum change in the full scale gain from the initial value using the internal 10 volt reference.

### APPLYING THE HI-574A

For each application of this converter, the ground connections, power supply bypassing, analog signal source, digital timing and signal routing on the circuit board must be optimized to assure maximum performance. These areas are reviewed in the following sections, along with basic operating modes and calibration requirements.

#### PHYSICAL MOUNTING AND LAYOUT CONSIDERATIONS

##### Layout –

Unwanted, parasitic circuit components, (L, R, and C) can make 12 bit accuracy impossible, even with a perfect A/D converter. The best policy is to eliminate or minimize these parasitics through proper circuit layout, rather than try to quantify their effects.

The recommended construction is a double-sided printed circuit board with a ground plane on the component side. Other techniques, such as wire-wrapping or point-to-point wiring on vectorboard, will have an unpredictable effect on accuracy.

In general, sensitive analog signals should be routed between ground traces and kept well away from digital lines. If analog and digital lines must cross, they should do so at right angles.

##### Power Supplies

Supply voltages to the HI-574A (+15V, -15V and +5V) must be "quiet" and well regulated. Voltage spikes on these lines can affect

#### TEMPERATURE COEFFICIENTS

The temperature coefficients for full-scale calibration, unipolar offset, and bipolar offset specify the maximum change from the initial (25°C) value to the value at  $T_{min}$  or  $T_{max}$ .

#### POWER SUPPLY REJECTION

The standard specifications for the HI-574A assume use of +5.00 and  $\pm 15.00$  or  $\pm 12.00$  volt supplies. The only effect of power supply error on the performance of the device will be a small change in the full scale calibration. This will result in a linear change in all lower order codes. The specifications show the maximum change in calibration from the initial value with the supplies at the various limits.

#### CODE WIDTH

A fundamental quantity for A/D converter specifications is the code width. This is defined as the range of analog input values for which a given digital output code will occur. The nominal value of a code width is equivalent to 1 least significant bit (LSB) of the full scale range or 2.44mV out of 10 volts for a 12-bit ADC.

#### QUANTIZATION UNCERTAINTY

Analog-to-digital converters exhibit an inherent quantization uncertainty of  $\pm \frac{1}{2}$ LSB. This uncertainty is a fundamental characteristic of the quantization process and cannot be reduced for a converter of given resolution.

#### LEFT-JUSTIFIED DATA

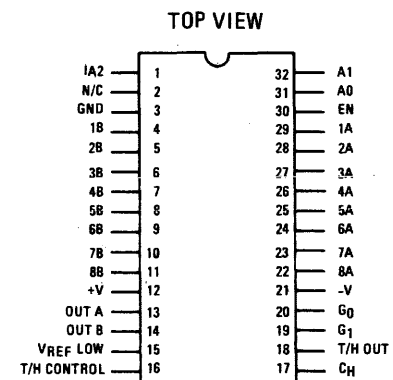
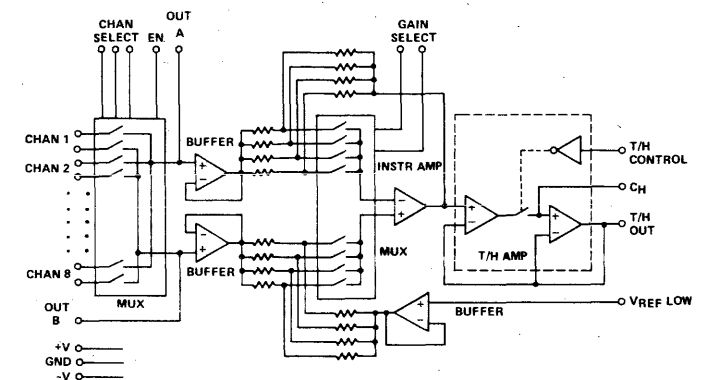
The data format used in the HI-574A is left-justified. This means that the data represents the analog input as a fraction of full-scale, ranging from 0 to  $\frac{4095}{4096}$ . This implies a binary point to the left of the MSB.

the converter's accuracy, causing several LSB's to flicker when a constant input is applied. Digital noise and spikes from a switching power supply are especially troublesome. If switching supplies must be used, outputs should be carefully filtered to assure "quiet" DC voltage at the converter terminals.

Further, a bypass capacitor pair on each supply voltage terminal is necessary to counter the effect of variations in supply current. Connect one pair from pin 1 to 15 ( $V_{Logic}$  supply), one from pin 7 to 9 ( $V_{CC}$  to Analog Common) and one from pin 11 to 9 ( $V_{EE}$  to Analog Common). For each capacitor pair, a  $10\mu F$  tantalum type in parallel with a  $0.1\mu F$  ceramic type is recommended.

#### Ground Connections

The typical HI-574A ground currents are 5.5mADC into pin 9 (Analog Ground) and 7mADC out of pin 15 (Digital Common). These pins should be tied together at the package to guarantee specified performance for the converter. In addition, a wide PC trace should run directly from pin 9 to (usually) 15V common, and from pin 15 to (usually) the +5V Logic Common. If the converter is located some distance from the system's "single point" ground, make only these connections to pins 9 and 15: Tie them together at the package, and back to the system ground with a single path. This path should have low resistance since it will carry about 1.5mA of DC current. (Code dependent currents flow in the  $V_{CC}$ ,  $V_{EE}$  and  $V_{Logic}$  terminals, but not through the HI-574A's Analog Common or Digital Common).

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>● INPUT OVERVOLTAGE PROTECTION</li> <li>● 50kHz THROUGHPUT</li> <li>● 12-BIT ACCURACY</li> <li>● OUTPUT TRACK/HOLD AMPLIFIER</li> <li>● ZERO OFFSET ADJUSTMENT</li> <li>● DIFFERENTIAL INPUT CHANNELS</li> <li>● SOFTWARE CONTROLLED GAIN AND CHANNEL SELECT</li> <li>● 85dB CMRR</li> <li>● COMPACT 32 PIN DIP</li> <li>● MIL-STD-883 SCREENING AVAILABLE</li> </ul>	<p>The HI-5900 comprises "front end" components of a data acquisition system including an eight channel differential multiplexer, programmable gain instrumentation amplifier (PGA), and Track and Hold amplifier. Adding a timing circuit and one A to D converter yields a complete data acquisition system. A 50kHz channel-to-channel throughput rate is achieved when the HI-5900 is used with a fast 12 bit A to D converter such as HARRIS HI-5712.</p> <p>Each output line of the input multiplexer is buffered by a high-quality non-inverting amplifier. This isolates each line from source resistances external to the 5900, preserving the high CMRR of the instrumentation amplifier block. Also, the buffers provide a high input impedance for each channel.</p> <p>The PGA, which includes an op amp, a monolithic resistor network and a four channel differential multiplexer, offers precision inverting gain values of -1, -2, -4, and -8. The voltage gain is selected by a two bit digital word. The output of the PGA drives the Track and Hold amplifier, and the ground side of the PGA is isolated by a buffer amplifier to maintain a high CMRR.</p> <p>The output Track/Hold amplifier is a monolithic device, internally connected for non-inverting unity gain. In the sample mode it operates as a high performance buffer amplifier. With an external holding capacitor, it may be switched to HOLD with an effective aperture delay time of 30ns.</p> <p>The packaging technique involves monolithic chips mounted in leadless chip carriers (LCC's) and soldered to both sides of a multilayer ceramic substrate. Each LCC may undergo reliability screening such as MIL-STD-883, Method 5004/Class B, before assembly on the substrate. The resulting package is a compact 32 pin DIP.</p> <p>The HI-5900 is offered as a high performance front-end section for military and industrial data acquisition systems. It is designed for interface with computers and is well suited for high-rel applications.</p>
<h3>APPLICATIONS</h3> <ul style="list-style-type: none"> <li>● HIGH PERFORMANCE DATA ACQUISITION</li> <li>● MILITARY SYSTEMS</li> </ul>	
<h3>PINOUT</h3> 	<h3>FUNCTIONAL DIAGRAM</h3> 

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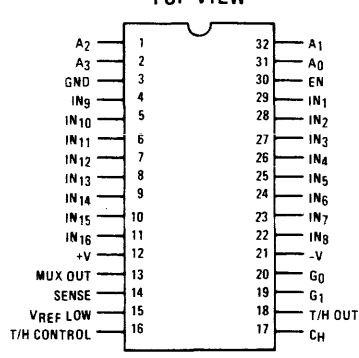
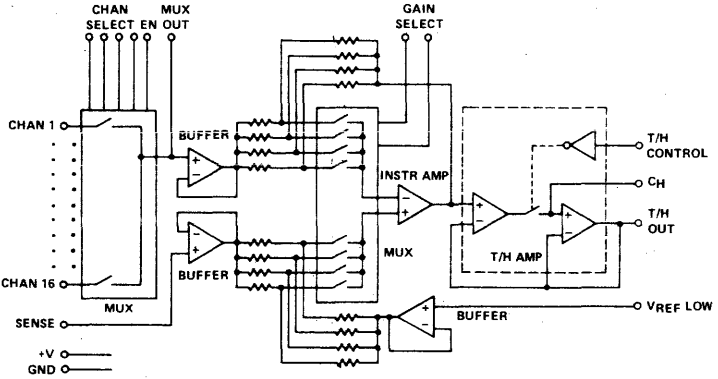
**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Voltage Between V+ and V- Terminals	40V	Output Current	Short Circuit Protected
Digital Input Overvoltage (Multiplexers)		Operating Temperature Range	
	V <sub>Supply</sub> (+) +4V	HA-5900-5	0°C ≤ T <sub>A</sub> ≤ +75°C
	V <sub>Supply</sub> (-) -4V	HA-5900-2	-55°C ≤ T <sub>A</sub> ≤ +125°C
Analog Input Overvoltage		Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C
	V <sub>Supply</sub> (+) +20V	Internal Power Dissipation	650mW
	V <sub>Supply</sub> (-) -20V	T/H Control Input	+8, -15V

**ELECTRICAL CHARACTERISTICS** Unless otherwise specified: V<sub>S</sub> = ±15V; C<sub>H</sub> = 1000pF; V<sub>IH</sub> = 4.0V; V<sub>IL</sub> = 0.8V

PARAMETER	TEMP	HI-5900-2 -55°C to +125°C			HI-5900-5 0°C to +70°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>ANALOG INPUT CHAR. EACH CHANNEL</b>								
Offset Voltage	+25°C		2	7		3	10	mV
	Full			9			12	mV
Bias Current	+25°C		80	300		80	300	nA
	Full		90	600		80	600	nA
Offset Current	+25°C		15	150		20	150	nA
	Full		30	300		30	300	nA
Common Mode Range	Full	±10			±10			V
Common Mode Rejection Ratio (V <sub>CM</sub> = ±10V) Any Gain	Full	80	85		74	85		dB
<b>DIGITAL INPUT CHAR.</b>								
Multiplexer Digital Input Current (High or Low)	Full		0.5	1		0.5	1	μA
Track/Hold Digital Input Current V <sub>IN</sub> ≤ 0.8V	Full			0.8			0.8	mA
V <sub>IN</sub> ≥ 4.0V	Full			20			20	μA
<b>TRANSFER CHARACTERISTICS</b>								
Small Signal Bandwidth (Gain = 1)	+25°C		2			2		MHz
Full Power Bandwidth (Gain = 1, V <sub>O</sub> = ±10V)	+25°C		70			70		kHz
Crosstalk (Sample Mode, Gain = 8, 1kHz 20VP -P Input on all but Selected Channel)	+25°C	-80	-90		-80	-90		dB
"Off Isolation (Hold Mode, Gain = 1, 1kHz 20V P-P Input)	+25°C		-76			-76		dB
Acquisition Time (Note 2), to 0.01% Gain - Absolute Error	+25°C		9			9		μs
Inverting Gain Of -1, -2,	Full		0.01	0.1		0.01	0.2	%
Inverting Gain Of -4, -8	Full		0.01	0.2		0.01	0.2	%
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing	Full	±10			±10			V
Output Current	+25°C	±10			±10			mA
Output Resistance	+25°C		5			5		Ω
<b>DYNAMIC CHARACTERISTICS</b>								
t <sub>ON</sub> , Enable (MUX)	+25°C		300			300		ns
t <sub>OFF</sub> , Enable (MUX)	+25°C		300			300		ns
Slew Rate	+25°C		±4			±4		V/μs
Droop Rate (T/H)	+25°C		5			5		nV/μs
	Full			20			5	μV/μs
Charge Transfer (T/H)	25°C		10			10		pC
Effective Aperture Delay Time (T/H)	+25°C		30			30		ns
Aperture Uncertainty (T/H)	+25°C		5			5		ns
<b>POWER SUPPLY CHARACTERISTICS</b>								
I <sub>+</sub>	+25°C			13			13	mA
	Full		8.5	15		8.0	15	mA
I <sub>-</sub>	+25°C			13			13	mA
	Full		6.5	15		6.0	15	mA
Power Supply Rejection Ratio, V+	Full	76	90		70	90		dB
Power Supply Rejection Ratio, V-	Full	80	100		80	100		dB

NOTES: 1. Absolute maximum ratings are limiting values, applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.  
 2. Acquisition Time is defined for a change of channel (+10V on chan. 1 to 0V on chan. 8) with simultaneous change from HOLD to TRACK mode. Gain = -1.

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>● INPUT OVERVOLTAGE PROTECTION</li> <li>● SOFTWARE CONTROLLED GAIN AND INPUT CHANNEL SELECTION</li> <li>● 16 PSEUDO-DIFFERENTIAL/SINGLE ENDED INPUT CHANNELS</li> <li>● INVERTING GAINS OF -1, -2, -4 AND -8</li> <li>● -90dB CROSSTALK</li> <li>● 0.01% GAIN ERROR</li> <li>● 9<math>\mu</math>s ACQUISITION TIME</li> <li>● DROOP RATE: 5nV/<math>\mu</math>sec</li> <li>● LOW POWER DISSIPATION 250mW</li> <li>● COMPACT 32 PIN DIP</li> <li>● MIL-STD-883 SCREENING AVAILABLE</li> </ul>	<p>The HI-5901 is a data acquisition front end subsystem intended for multisensor based high-level applications, requiring conversion of analog input data to digital form for computer processing. It provides sixteen single-ended or pseudo-differential channels of fault-protected multiplexed inputs, programmable gains of -1, -2, -4, -8 and a buffered track and hold output block compatible with any commercially available A/D converter. All these functions are digitally selectable through appropriate coding of seven control terminals. Input channel expansion can be easily implemented through addition of external multiplexers and proper utilization of the enable-command pin.</p> <p>Being self-contained units except for the holding capacitor, they facilitate user applications and eliminate the need for selection of high-priced precision resistors or labor intensive adjustments to achieve the accuracy levels specified.</p> <p>This product provides channel to channel throughput rates of 50kHz at <math>\pm 10</math> volt signal range when used in connection with a fast 12 bit A/D converter such as the HI-5712. In addition, it offers excellent input characteristics such as low input offset voltage with offset nulling capability, low input currents, very high input impedance, and very low crosstalk. Typical acquisition time and gain error are 9 microseconds and <math>\pm 0.01\%</math>, respectively. The internal track and hold amplifier features an effective aperture delay time of 30ns and a droop rate of 5nV/<math>\mu</math>sec. Total power dissipation is only 250mW.</p>
<h3>APPLICATIONS</h3> <ul style="list-style-type: none"> <li>● MULTI-CHANNEL DATA ACQUISITION SYSTEMS</li> <li>● STATUS MONITORING SYSTEMS</li> <li>● PROCESS CONTROL SYSTEMS</li> <li>● INSTRUMENTATION</li> <li>● HIGH RELIABILITY DAS's</li> </ul>	<p>A complete high-speed and high precision data acquisition system with 15 bits of dynamic range can be easily implemented with only three components: the HI-5901, the HI-5712, and an offset nulling DAC. Board space required is 3 square inches and total weight is less than 25 grams.</p> <p>The manufacturing technique adopted for the HI-5901 involves monolithic dice packaged in leadless chip carriers (LCC's) and soldered to both sides of a multilayer ceramic substrate. The resulting product is a compact and easy-to-use 32 pin DIP.</p> <p>The HI-5901 is intended for military, aerospace, industrial and instrumentation applications. MIL-STD-883 Class B and high reliability commercial grades are both available as standard products.</p>
<h3>PINOUT</h3> <p style="text-align: center;">TOP VIEW</p> 	<h3>FUNCTIONAL DIAGRAM</h3> 

Copyright © Harris Corporation 1983



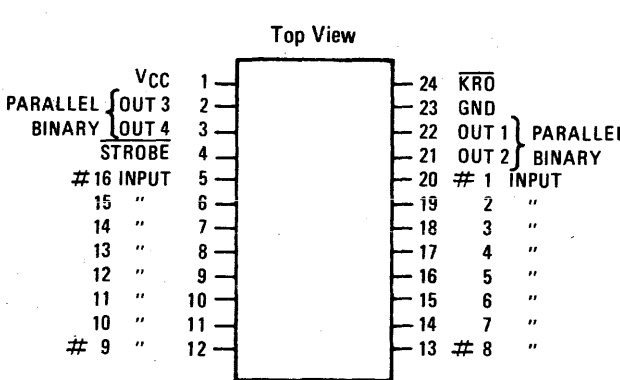
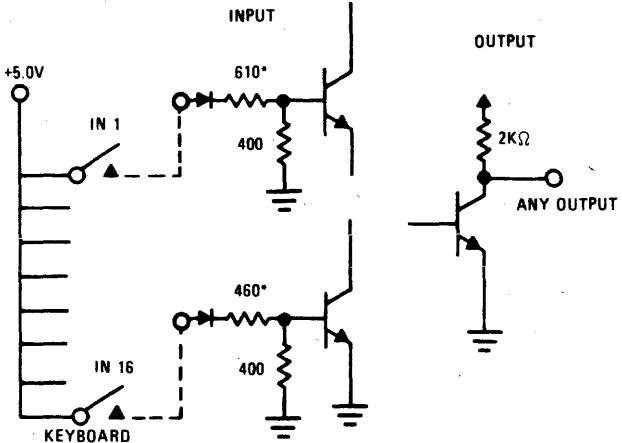
**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Voltage Between V+ and V- Terminals	40V	Output Current	Short Circuit Protected
Digital Input Overvoltage (Multiplexers)	V <sub>Supply</sub> (+) +4V V <sub>Supply</sub> (-) -4V	Operating Temperature Range	0°C ≤ T <sub>A</sub> ≤ +75°C -55°C ≤ T <sub>A</sub> ≤ +125°C
Analog Input Overvoltage	V <sub>Supply</sub> (+) +20V V <sub>Supply</sub> (-) -20V	Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C
		Internal Power Dissipation	650mW
		T/H Control Input	+8, -15V

**ELECTRICAL CHARACTERISTICS** Unless otherwise specified: V<sub>S</sub> = ±15V; C<sub>H</sub> = 1000pF; V<sub>IH</sub> = 4.0V; V<sub>IL</sub> = 0.8V

PARAMETER	TEMP	HI-5901-2, -8 -55°C to +125°C			HI-5901-5, -7 0°C to +70°C			UNITS
		MIN	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>ANALOG INPUT CHAR. EACH CHANNEL</b>								
Offset Voltage	+25°C Full		2 7.5	9.5		3 10.5	13	mV mV
Bias Current	+25°C Full		80 90	300 600		80 80	300 600	nA nA
Offset Current	+25°C Full		15 30	150 300		20 30	150 300	nA nA
Common Mode Range	Full	±10			±10			V
Common Mode Rejection Ratio (V <sub>CM</sub> = ±10V) Any Gain	Full	80	85		74	85		dB
<b>DIGITAL INPUT CHAR.</b>								
Multiplexer Digital Input Current (High or Low)	Full		0.5	1		0.5	1	μA
Track/Hold Digital Input Current V <sub>IN</sub> ≤ 0.8V	Full			0.8			0.8	mA
V <sub>IN</sub> ≥ 4.0V	Full			20			20	μA
<b>TRANSFER CHARACTERISTICS</b>								
Small Signal Bandwidth (Gain = 1)	+25°C		2			2		MHz
Full Power Bandwidth (Gain = 1, V <sub>O</sub> = ±10V)	+25°C		70			70		kHz
Crosstalk (Sample Mode, Gain = 8, 1kHz 20VP -P Input on all but Selected Channel)	+25°C	-80	-90		-80	-90		dB
Off Isolation (Hold Mode, Gain = 1, 1kHz 20V P-P Input)	+25°C		-76			-76		dB
Acquisition Time (Note 2), to 0.01% Gain - Absolute Error	+25°C		9			9		μs
Inverting Gain Of -1, -2, Inverting Gain Of -4, -8	Full		0.01 0.01	0.1 0.2		0.01 0.01	0.2 0.2	% %
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing	Full	±10			±10			V
Output Current	+25°C	±10			±10			mA
Output Resistance	+25°C		5			5		Ω
<b>DYNAMIC CHARACTERISTICS</b>								
t <sub>ON</sub> , Enable (MUX)	+25°C		300			300		ns
t <sub>OFF</sub> , Enable (MUX)	+25°C		300			300		ns
Slew Rate	+25°C		±4			±4		V/μs
Droop Rate (T/H)	+25°C Full		5	20		5	5	nV/μs μV/μs
Charge Transfer (T/H)	25°C		10			10		pC
Effective Aperture Delay Time (T/H)	+25°C		30			30		ns
Aperture Uncertainty (T/H)	+25°C		5			5		ns
<b>POWER SUPPLY CHARACTERISTICS</b>								
I <sub>+</sub>	+25°C Full		8.5	13 15		8.0	13 15	mA mA
I <sub>-</sub>	+25°C Full		6.5	13 15		6.0	13 15	mA mA
Power Supply Rejection Ratio, V+	Full	76	90		70	90		dB
Power Supply Rejection Ratio, V-	Full	80	100		80	100		dB

NOTES: 1. Absolute maximum ratings are limiting values, applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.  
2. Acquisition Time is defined for a change of channel (+10V on chan. 1 to 0V on chan.16) with simultaneous change from HOLD to TRACK mode. Gain = -1.

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• STROBE OUTPUT</li> <li>• KEY ROLLOVER OUTPUT</li> <li>• EXPANDABLE: 2 PACKAGES REQUIRED FOR FULL TELETYPEWRITER, EIGHT-BIT ENCODING</li> <li>• SINGLE +5.0V SUPPLY REQUIRED</li> <li>• DTL/TTL OUTPUTS</li> <li>• MONOLITHIC RELIABILITY</li> </ul>	<p>The HD-0165 Keyboard Encoder is a 16 line to four-bit parallel encoder intended for use with manual data entry devices such as calculator or typewriter keyboards. In addition to the encoding function, there is a Strobe output and a Key Rollover output which energizes whenever two or more inputs are energized simultaneously. Any four-bit code can be implemented by proper wiring of the input lines. Inputs are normally wired through the key switches to the +5.0V power supply. Full typewriter keyboard encoding up to eight bits can be accomplished with two Encoder circuits by the use of double pole key switches or single pole switches with two isolation diodes per key. Outputs will interface with all popular DTL and TTL logic families. The circuit is packaged in a hermetic 24-pin dual-in-line package and operates over the temperature range of 0°C to +75°C.</p>
<h3>APPLICATIONS</h3> <ul style="list-style-type: none"> <li>• MICROPROCESSOR DATA ENTRY (16 KEY TO HEX CODE)</li> <li>• BCD DATA ENTRY</li> <li>• TYPEWRITER TYPE KEYBOARDS</li> <li>• CONTROL PANELS</li> </ul>	
<h3>PINOUT</h3>	<h3>EQUIVALENT CIRCUITS</h3>
 <p style="text-align: center;">Top View</p>	

\* EQUIVALENT RESISTORS FOR OTHER INPUTS ARE BETWEEN THESE TWO VALUES

**SPECIFICATIONS**



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage +7.0V  
 Input Voltage +5.5V  
 Output Voltage +5.5V

Output Current 30mA  
 Storage Temperature -65° to +150°C  
 Operating Temperature (Case) 0°C to +75°C

**ELECTRICAL CHARACTERISTICS**

Test Conditions:  $V_{CC} = +5.0V \pm 5\%$   
 $T_{Case} = 0^{\circ}C \text{ to } +75^{\circ}C$   
 Unless otherwise specified

PARAMETER	SYM.	LIMITS			UNITS	TEST CONDITIONS
		MIN.	TYP.	MAX.		
Input Current	"1" I <sub>IH</sub>			17	mA	V <sub>IN</sub> = +5.0V
Output Voltage	"0" V <sub>OL</sub>		+0.2	+0.4	V	V <sub>IH</sub> = +4.5V I <sub>OL</sub> = 10mA V <sub>IH</sub> = +3.75V I <sub>OL</sub> = 3.2mA V <sub>IL</sub> = Open Circuit, I <sub>OH</sub> = -240µA
	"1" V <sub>OH</sub>	+2.4	+4.0			
Power Supply Current	Operating I <sub>CC</sub>			52	mA	One Input at +5.25V
	Maximum I <sub>CCM</sub>			88	mA	All Inputs at +5.25V
A.C. Skew Time (Note 1)	T <sub>SK</sub>		80	200	ns	T <sub>Case</sub> = 25°C V <sub>CC</sub> = V <sub>IN</sub> = +5.0V C <sub>L</sub> < 50pF

NOTE: (1) Skew time is the maximum time differential between propagation delay times of any outputs including strobe and K<sub>RO</sub>.

**TRUTH TABLE**

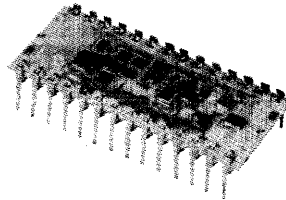
INPUTS																OUTPUTS					
1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	1	2	3	4	St.	K <sub>RO</sub>
L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	H	H	H	H	H	H
H	L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	H	H	H	H	L	H
L	H	L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	H	H	H	L	H
L	L	H	L	L	L	L	L	L	L	L	L	L	L	L	L	H	L	H	H	L	H
L	L	L	H	L	L	L	L	L	L	L	L	L	L	L	L	L	L	H	H	L	H
L	L	L	L	H	L	L	L	L	L	L	L	L	L	L	L	H	H	L	H	L	H
L	L	L	L	L	H	L	L	L	L	L	L	L	L	L	L	H	L	L	H	L	H
L	L	L	L	L	L	H	L	L	L	L	L	L	L	L	L	L	L	L	H	L	H
L	L	L	L	L	L	L	H	L	L	L	L	L	L	L	L	H	H	H	L	L	H
L	L	L	L	L	L	L	L	H	L	L	L	L	L	L	L	L	H	H	L	L	H
L	L	L	L	L	L	L	L	L	H	L	L	L	L	L	L	H	L	H	L	L	H
L	L	L	L	L	L	L	L	L	L	H	L	L	L	L	L	L	L	H	L	L	H
L	L	L	L	L	L	L	L	L	L	L	H	L	L	L	L	L	L	H	L	L	H
L	L	L	L	L	L	L	L	L	L	L	L	H	L	L	L	H	L	L	L	L	H
L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	H	L	L	L	L	L	H
L	L	L	L	L	L	L	L	L	L	L	L	L	L	L	H	L	L	L	L	L	H
ANY TWO OR MORE HIGH																X	X	X	X	L	L

INPUTS: L = Open Circuit or < +1.0V H = > +4.5V Current Source  
 OUTPUTS: L = < +0.4V H = > +2.4V X = Erroneous Data



# 8 Channel, 12 Bit DAS with $\mu$ P Interface

HS 9410 Series



**\$99 (100's)**

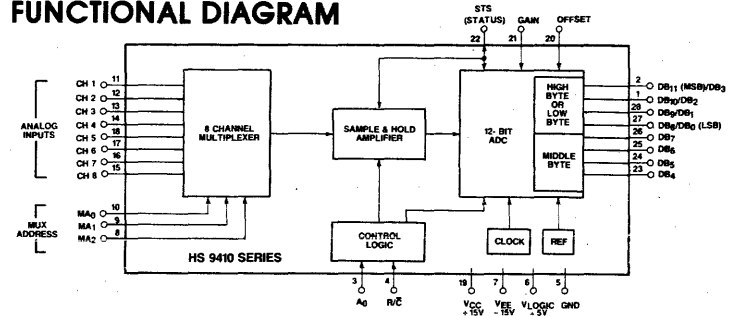
## FEATURES

- Complete 8 Channel, 12-Bit Data Acquisition System with MUX, SH, REF, Clock and three-state outputs
- Full 8- or 16-Bit Microprocessor Bus Interface
- Guaranteed Linearity Over Temperature
- High Throughput Rate: 30kHz
- Hermetic 28-pin Metal or Low Cost Epoxy DIP
- Low Power: 900mW

## SPECIFICATIONS

Resolution	12 Bits
Number of Channels	8 Single Ended
Throughput Rate	30 kHz
Power Supply	$\pm 15V$ , +5V
Package	28-pin DIP
Other Specifications	See Ordering Guide

## FUNCTIONAL DIAGRAM



## ORDERING GUIDE

MODEL NUMBER <sup>1</sup>	INPUT RANGE	SYSTEM ACCURACY (% FSR)	FULL SCALE T.C. (PPM/ $^{\circ}$ C)	TEMP. RANGE	SCREENING	PRICE (1-9)
HS 94XXJ	SEE NOTE 1	$\pm 0.025$	50.0	0 $^{\circ}$ C to +70 $^{\circ}$ C	0.4% AQL*	\$115.00
HS 94XXX		$\pm 0.012$	20.0	0 $^{\circ}$ C to +70 $^{\circ}$ C	0.4% AQL*	125.00
HS 94XXS		$\pm 0.025$	50.0	-55 $^{\circ}$ C to +85 $^{\circ}$ C	0.4% AQL*	215.00
HS 94XXT		$\pm 0.012$	20.0	-55 $^{\circ}$ C to +85 $^{\circ}$ C	0.4% AQL*	265.00
HS 94XXS/B		$\pm 0.025$	50.0	-55 $^{\circ}$ C to +85 $^{\circ}$ C	MIL-STD-883B	370.00
HS 94XXT/B		$\pm 0.012$	20.0	-55 $^{\circ}$ C to +85 $^{\circ}$ C	MIL-STD-883B	450.00

NOTES:

1. MODEL HS 94XX

MODEL SUFFIX	INPUT RANGE
10	0 to +10V
11	$\pm 5V$
12	$\pm 10V$

\*0.4% AQL All units have full burn-in at +85 $^{\circ}$ C. Acceptable quality level (AQL) of 0.4% means that Hybrid Systems guarantees that there will be no rejects in a sample lot of 100 pieces. That's more than twice as tough as it has to be - even for military applications. Specifications subject to change without notice.

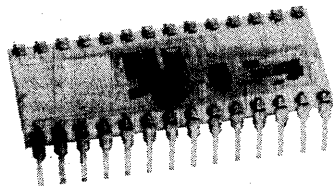
Add letter suffix as required above.

## Technical knockouts from the Contenders!

# Industry Standard 574

12-Bit, 30 $\mu$ sec Hybrid ADC

**\$34.50**  
**(100's)**



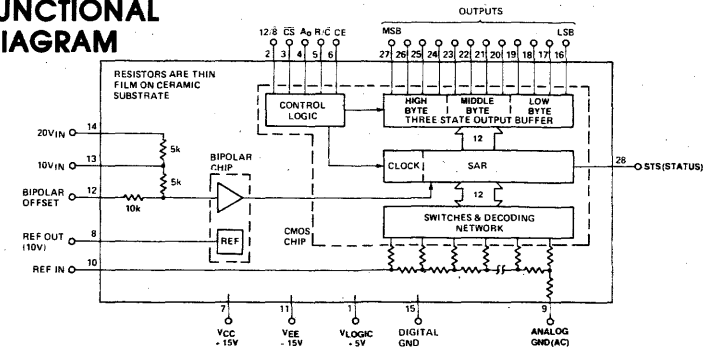
## FEATURES

- Complete 12-Bit A/D Converter
- 30 $\mu$ Sec Conversion Time
- Power 400mW
- No Missing Codes
- $\mu$ P BUS Compatible
- Commercial and MIL-STD-883B Processing
- Hermetic, 28-pin DIP, Metal Package

## SPECIFICATIONS

Resolution	12 Bits
Linearity Error	No Missing Codes
Differential Integral	$\pm 0.013\%$ F.S.R.
End Point Error	
Offset Bipolar	$\pm 0.1\%$ F.S.R.
Unipolar	$\pm 0.05\%$ F.S.R.
Full Scale	$\pm 0.3\%$ F.S.R.
Conversion Time	30 $\mu$ sec
Temperature Stability	
Linearity	$\pm 0.5$ ppm/ $^{\circ}$ C F.S.R.
Offset	$\pm 5$ ppm/ $^{\circ}$ C F.S.R.
Full Scale	$\pm 10$ ppm/ $^{\circ}$ C F.S.R.
Power	$\pm 15V$ , +5V/600mW
Package	28-pin DIP

## FUNCTIONAL DIAGRAM



## ORDERING GUIDE

MODEL NUMBER	RESOLUTION NO MISSING CODES (T <sub>min</sub> to T <sub>max</sub> )	LINEARITY ERROR	TEMP. RANGE	SCREENING	PRICE (1-9)
HS 574J	11 Bits	$\pm 1$ LSB	0 $^{\circ}$ C to +70 $^{\circ}$ C	0.4% AQL*	49.50
HS 574K	12 Bits	$\pm 1/2$ LSB	0 $^{\circ}$ C to +70 $^{\circ}$ C	0.4% AQL*	61.50
HS 574L	12 Bits	$\pm 1/2$ LSB	0 $^{\circ}$ C to +70 $^{\circ}$ C	0.4% AQL*	89.00
HS 574S	11 Bits	$\pm 1$ LSB	-55 $^{\circ}$ C to +125 $^{\circ}$ C	0.4% AQL*	136.00
HS 574T	12 Bits	$\pm 1$ LSB'	-55 $^{\circ}$ C to +125 $^{\circ}$ C	0.4% AQL*	183.00
HS 574U	12 Bits	$\pm 1$ LSB'	-55 $^{\circ}$ C to +125 $^{\circ}$ C	0.4% AQL*	265.00
HS 574S/B	11 Bits	$\pm 1$ LSB'	-55 $^{\circ}$ C to +125 $^{\circ}$ C	883B	164.00
HS 574T/B	12 Bits	$\pm 1$ LSB'	-55 $^{\circ}$ C to +125 $^{\circ}$ C	883B	220.00
HS 574U/B	12 Bits	$\pm 1$ LSB'	-55 $^{\circ}$ C to +125 $^{\circ}$ C	883B	315.00

\*Note:  $\pm 1/2$  LSB max error from -25 $^{\circ}$ C to -85 $^{\circ}$ C

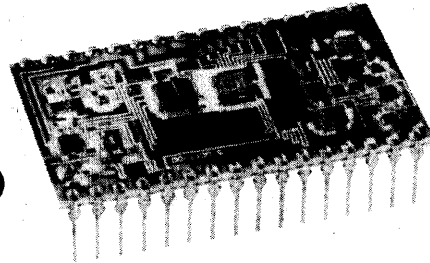
# Complete 0.0008% 16-Bit ADC

HS 9516 Series

## FEATURES

- True 16-bit ( $\pm 0.00081$ ) Linearity Error
- Full 16-bit Resolution (1 part in 65,536)
- No Missing Codes (16-bits) 0°C to +70°C
- Six User Selectable Input Ranges
- 100 $\mu$ sec Conversion Time
- Low Full Scale Drift 10ppm/°C (Max)
- Low Power - 90mW Typical

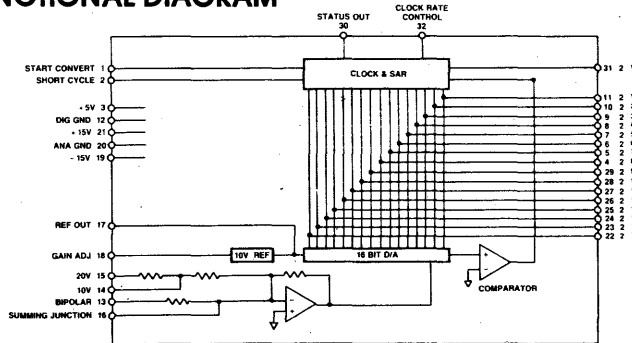
**\$239.00 (1-9)**



## SPECIFICATIONS

Resolution	16 Bits
Analog Inputs	0 to +5V, 0 to +10V, 0 to +20V
Unipolar	$\pm 2.5V, \pm 5V, \pm 10V$
Bipolar	Parallel, OBITN, 2's Comp
Outputs	
Accuracy	
Scale Factor Error	$\pm 0.1\%$ max (Adj. to Zero)
Zero Error	$\pm 0.1\%$ FSR max (Adj. to Zero)
(Unipolar or Bipolar)	
Differential Linearity Error	
HS 9516-6	$\pm 0.0015\%$ FSR ( $\pm 1$ LSB) max
HS 9516-5	$\pm 0.003\%$ FSR ( $\pm 2$ LSB) max
HS 9516-4	$\pm 0.006\%$ FSR ( $\pm 4$ LSB) max
No Missing Codes	0°C to +70°C
Integral Linearity Error	
HS 9516-6	$\pm 0.0008\%$ FSR ( $\pm 1/2$ LSB) max
HS 9516-5	$\pm 0.0015\%$ FSR ( $\pm 1$ LSB) max
HS 9516-4	$\pm 0.003\%$ FSR ( $\pm 2$ LSB) max
Stability	
Scale Factor Tempco	$\pm 10$ ppm/°C max
Zero Tempco	
Unipolar	$\pm 2$ ppm FSR/°C
Bipolar	$\pm 5$ ppm FSR/°C
Integral Linearity Tempco	$\pm 1$ ppm FSR/°C max
Differential Linearity	$\pm 0.5$ ppm FSR/°C max
Power	$\pm 15V, +5V$
Package	32-pin Triple DIP
Temperature Range	0°C to +70°C

## FUNCTIONAL DIAGRAM



## ORDERING GUIDE

MODEL	DESCRIPTION	PRICE
HS 9516-4	16-Bit ADC with 14-Bit Integral & Differential Linearity	\$239.00
HS 9516-5	16-Bit ADC with 15-Bit Integral & Differential Linearity	249.00
HS 9516-6	16-Bit ADC with 16-Bit Integral & Differential Linearity	299.00

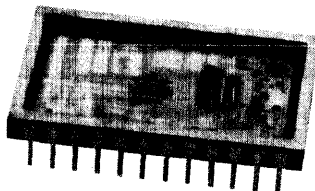
**Technical knockouts from the Contenders!**

# 12-Bit Adjustment Free ADC's

HS 5200 Series

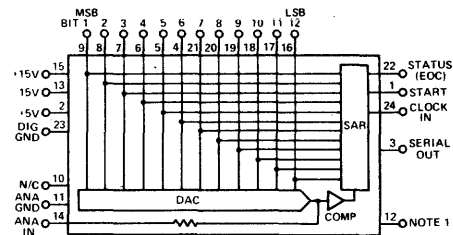
## FEATURES

- 12-Bit Conversion in 10 $\mu$ Sec (Typical)
- Adjustment-Free ( $\pm 0.0125\%$ ) Linearity
- Low Power 670mW (Typical)
- Hermetic, 24-pin DIP, Metal Package
- MIL-STD-883B Processing
- Pin Compatible to MN5200



**\$160.00 (1-9)**

## FUNCTIONAL DIAGRAM



## ORDERING GUIDE

MODEL	RESOLUTION / CONVERSION TIME	INPUT RANGE	TEMP RANGE	PROCESSING	PRICE (1-9)
HS 5210C-12	12 Bits, 13 $\mu$ Sec	0 to -10V	0°C to +70°C	0.4% AQL*	\$160.00
HS 5211C-12	12 Bits, 13 $\mu$ Sec	$\pm 5V$	0°C to +70°C	0.4% AQL*	\$160.00
HS 5212C-12	12 Bits, 13 $\mu$ Sec	$\pm 10V$	0°C to +70°C	0.4% AQL*	\$160.00
HS 5213C-12	12 Bits, 13 $\mu$ Sec	0 to -10V	0°C to +70°C	0.4% AQL*	\$160.00
HS 5214C-12	12 Bits, 13 $\mu$ Sec	$\pm 5V$	0°C to +70°C	0.4% AQL*	\$160.00
HS 5215C-12	12 Bits, 13 $\mu$ Sec	$\pm 10V$	0°C to +70°C	0.4% AQL*	\$160.00
HS 5216C-12	12 Bits, 13 $\mu$ Sec	0 to +10V	0°C to +70°C	0.4% AQL*	\$160.00
HS 5210B-12	12 Bits, 13 $\mu$ Sec	0 to -10V	-55°C to +125°C	MIL-STD-883B	\$190.00
HS 5211B-12	12 Bits, 13 $\mu$ Sec	$\pm 5V$	-55°C to +125°C	MIL-STD-883B	\$190.00
HS 5212B-12	12 Bits, 13 $\mu$ Sec	$\pm 10V$	-55°C to +125°C	MIL-STD-883B	\$190.00
HS 5213B-12	12 Bits, 13 $\mu$ Sec	0 to -10V	-55°C to +125°C	MIL-STD-883B	\$190.00
HS 5214B-12	12 Bits, 13 $\mu$ Sec	$\pm 5V$	-55°C to +125°C	MIL-STD-883B	\$190.00
HS 5215B-12	12 Bits, 13 $\mu$ Sec	$\pm 10V$	-55°C to +125°C	MIL-STD-883B	\$190.00
HS 5216B-12	12 Bits, 13 $\mu$ Sec	0 to +10V	-55°C to +125°C	MIL-STD-883B	\$190.00

## SPECIFICATIONS

Resolution	12 Bits
Linearity	$\pm 1/2$ LSB
Absolute Accuracy <sup>1</sup>	$\pm 0.4\%$
Conversion Time	13 $\mu$ sec
Temperature Range	
C-versions	0°C to +70°C
B-versions	-55°C to +125°C
Power	+15V, +5V/870mW
Package	24-pin DIP, Metal

<sup>1</sup>Includes all errors, gain, zero, and linearity over temperature.

**0.4% A.Q.L.\*** All units have full power burn-in at +85°C. Acceptable quality level (AQL) of 0.4% means that Hybrid Systems guarantees that there will be no rejects in a sample lot of 100 pieces. That's more than twice as tough as it has to be—even for military applications.

\*10V external reference user supplied

**Hybrid Systems**  
CORPORATION

22 Linnell Circle, Billerica, MA 01821 (617)667-8700, TWX 710-347-1575

# 12-Bit 10 $\mu$ Sec Hi Rel ADC

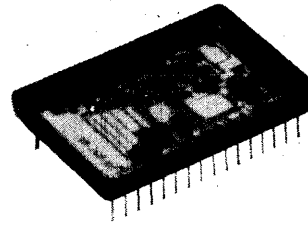
ADC85 Series

## FEATURES

- -55°C to +125°C Operation
- No Missing Codes
- Replaces ADC84/85, HX12B, HZ12B
- Low Power 1.2W (maximum)

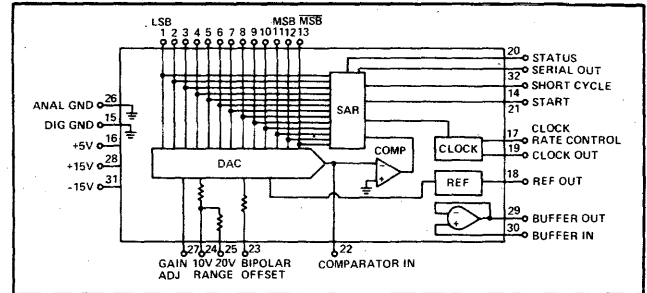
## SPECIFICATIONS

Resolution	12 Bits
Analog Inputs	
Unipolar	0 to +5V, 0 to +10V
Bipolar	$\pm 2.5V$ , $\pm 5V$ , $\pm 10V$
Outputs	Parallel, Serial/TTL
Coding	CBIN/COBIN
Linearity	$\pm 0.012\%$ F.S.R. (maximum)
Conversion Time	8.8 $\mu$ Sec, 10 $\mu$ Sec (maximum)
Scale Factor Drift	15ppm/ $^{\circ}C$ (maximum)
Linearity Drift	2ppm/ $^{\circ}C$ (maximum)
Temperature Range	-55 $^{\circ}C$ to +125 $^{\circ}C$
Power	$\pm 15V$ , +5V/1.2W (maximum)
Package	32-pin DIP



**\$134.00 (1-9)**

## FUNCTIONAL DIAGRAM



## ORDERING GUIDE

MODEL	RESOLUTION	ACCURACY	TEMP RANGE	PROCESSING	PRICE
HS ADC85B	12 Bits	12 Bits	-55 $^{\circ}C$ to +125 $^{\circ}C$	MIL-STD-883B	\$205.00
HS ADC85C	12 Bits	12 Bits	0 $^{\circ}C$ to +70 $^{\circ}C$	0.4% AQL*	\$134.00

**Technical knockouts from the Contenders!**

# Complete $\mu$ P Compatible 12 Bit DAC

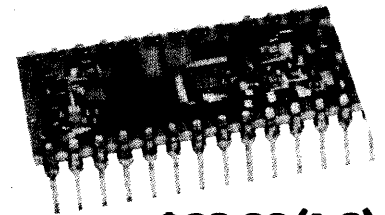
DAC338, HS 9338 Series

## FEATURES

- $\mu$ P BUS Compatible
- 0 to +10V, 0 to +5V,  $\pm 10V$ ,  $\pm 5V$  Output Ranges
- Binary Coding
- Linearity  $\pm 0.01\%$
- 2.5 $\mu$ Sec Settling Time

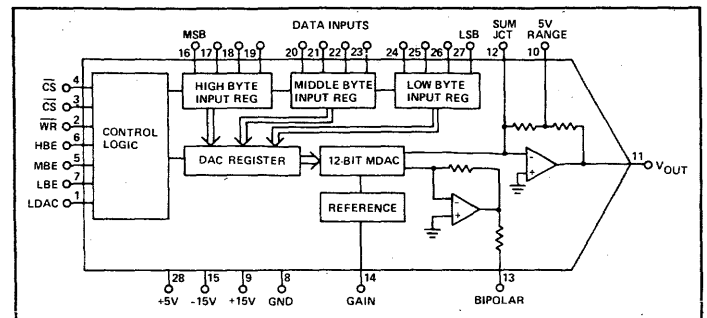
## SPECIFICATIONS

Resolution	12 Bits
Coding	BIN, OBIN
Logic	CMOS, TTL
Output	0 to +10V, $\pm 10V$ , $\pm 5V$
Linearity	
-0	$\pm 0.05\%$ F.S.R.
-1	$\pm 0.02\%$ F.S.R.
-2	$\pm 0.01\%$ F.S.R.
Settling Time	20 $\mu$ Sec
Scale Factor Drift	15ppm/ $^{\circ}C$
Temperature Range	0 $^{\circ}C$ to +70 $^{\circ}C$
Power	+15V/450mW
Package	28-pin DIP



**\$39.00 (1-9)**

## FUNCTIONAL DIAGRAM



## ORDERING GUIDE

MODEL	RESOLUTION	LINEARITY	TEMP RANGE	PROCESSING	PRICE (1-9)
DAC338B-12-0	12 Bits	10 Bits	-55 $^{\circ}C$ to +125 $^{\circ}C$	MIL-STD-883B	\$118.00
DAC338B-12-1	12 Bits	11 Bits	-55 $^{\circ}C$ to +125 $^{\circ}C$	MIL-STD-883B	\$126.00
DAC338B-12-2	12 Bits	12 Bits	-55 $^{\circ}C$ to +125 $^{\circ}C$	MIL-STD-883B	\$137.00
HS 9338-0	12 Bits	10 Bits	0 $^{\circ}C$ to +70 $^{\circ}C$	0.4% AQL*	\$ 39.00
HS 9338-1	12 Bits	11 Bits	0 $^{\circ}C$ to +70 $^{\circ}C$	0.4% AQL*	\$ 44.00
HS 9338-2	12 Bits	12 Bits	0 $^{\circ}C$ to +70 $^{\circ}C$	0.4% AQL*	\$ 49.00

**Hybrid Systems**

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# 14, 15 and 16 Bit Linearity Latched MDAC's

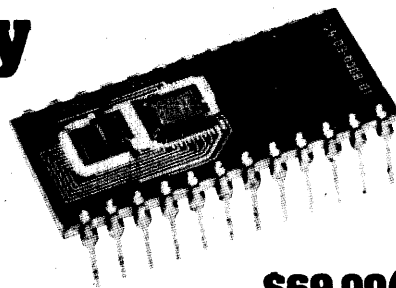
## DAC9331-16 Series

### FEATURES

- Up to 16-Bit (0.0008%) Linearity
- Two-Chip Construction
- Input Registers
- 24-pin DIP, Metal Package
- 2 and 4-Quadrant Multiplication
- Single Supply Operation
- Low Power 60mW

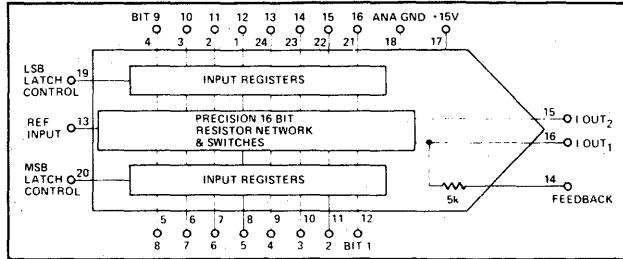
### SPECIFICATIONS

Resolution	16 Bits
Coding	BIN, OBIN
Logic	CMOS, TTL
Reference Input	0 to $\pm 25V$
Output	$200\mu A/V_{REF}$
Linearity	
DAC9331-16-4	$\pm 0.003\%$ F.S.R.
DAC9331-16-5	$\pm 0.0015\%$ F.S.R.
DAC9331-16-6	$\pm 0.0008\%$ F.S.R.
Settling Time	2 $\mu$ Sec
Scale Factor Drift	2ppm/ $^{\circ}C$
Temperature Range	$0^{\circ}C$ to $+70^{\circ}C$
Power	+15V/60mW
Package	24-pin DIP



**\$69.00(1-9)**

### FUNCTIONAL DIAGRAM



### ORDERING GUIDE

MODEL	RESOLUTION	LINEARITY	TEMP RANGE	PROCESSING	PRICE (1-9)
DAC9331-16-4	16 Bits	14 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$ 69.00
DAC9331-16-5	16 Bits	15 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$109.00
DAC9331-16-6	16 Bits	16 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$169.00

\*\$99.00 (100's)

## Technical knockouts from the Contenders!

# Complete Buffered 16 and 18 Bit DAC's

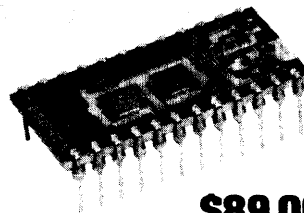
## DAC9377, 377 Series

### FEATURES

- True 16-Bit (0.0008%) Linearity
- $\mu P$  Compatible
- Complete DAC
- 24-pin DIP, Metal Package
- Low Power 450mW

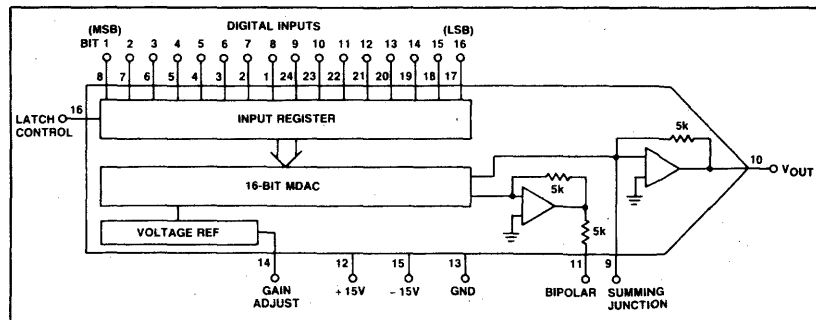
### SPECIFICATIONS

Resolution	16 Bits, 18 Bits
Coding	BIN, OBIN
Logic	CMOS, TTL
Output	0 to $+10V$ , $\pm 10V$
Linearity	
DAC377-18	$\pm 0.0008\%$ F.S.R.
DAC9337-16-6	$\pm 0.0008\%$ F.S.R.
DAC9337-16-5	$\pm 0.0015\%$ F.S.R.
DAC9337-16-4	$\pm 0.003\%$ F.S.R.
DAC9377-4D	$\pm 0.002\%$ F.S.R.
Settling Time	20 $\mu$ Sec
Scale Factor Drift	5ppm/ $^{\circ}C$
Temperature Range	$0^{\circ}C$ to $+70^{\circ}C$
Power	+15V/450mW
Package	24-pin DIP (9377) 28-pin DIP, Metal (377)



**\$89.00(1-9)**

### FUNCTIONAL DIAGRAM



### ORDERING GUIDE

MODEL	RESOLUTION	LINEARITY	TEMP RANGE	PROCESSING	PRICE (1-9)
DAC9377-4D	4 BCD	$\pm 0.002\%$	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$ 99.00
DAC9377-16-4	16 Bits	14 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$ 89.00
DAC9377-16-5	16 Bits	16 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$129.00
DAC9377-16-6	16 Bits	16 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$199.00
DAC377B-18	18 Bits	16 Bits	$-55^{\circ}C$ to $+125^{\circ}C$	MIL-STD-883B	\$395.00
DAC377C-18	18 Bits	16 Bits	$0^{\circ}C$ to $+70^{\circ}C$	0.4% AQL*	\$205.00

**Hybrid Systems**  
CORPORATION

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**DATA ACQUISITION**



**Integrating Analog-to-Digital Converters for Display**

Maximum Electrical Specification at 25°C unless otherwise noted.

Model	ICL7136	ICL7137	ICL7135	ICL7129	ICL7106/ICL7116	ICL7107/ICL7117
<b>Resolution</b>	± 3½ Digit	± 3½ Digit	± 4½ Digit	± 4½ Digit	± 3½ Digit	± 3½ Digit
<b>Accuracy</b>						
Non-Linearity	± 1 Count	± 1 Count	± 1 Count	± 1 Count	± 1 Count	± 1 Count
Zero Input Reading	± 0.000	± 0.000	± 0.000	± 0.000	± 0.000	± 0.000
Ratiometric Reading	± 1.000	± 1.000	± 1.000	± 0.9997	± 1.000	± 1.000
V <sub>IN</sub> = V <sub>REF</sub>	± 1 Count	± 1 Count	± 1 Count	± 3 Counts	± 1 Count	± 1 Count
Rollover Error	± 1 Count	± 1 Count	± 1 Count	± 1 Count	± 1 Count	± 1 Count
<b>Stability</b>						
Offset vs. Temperature	1µV/°C	1µV/°C	1µV/°C	1µV/°C	1µV/°C	1µV/°C
Gain vs. Temperature	5 ppm/°C	5 ppm/°C	5 ppm/°C	5 ppm/°C	5 ppm/°C	5 ppm/°C
<b>Conversion Time</b>	0.1 to 3 conv/sec	0.1 to 3 conv/sec	0.1 to 15 conv/sec	0.1 to 6 conv/sec	0.1 to 15 conv/sec	0.1 to 15 conv/sec
<b>Analog Input</b>						
Voltage Range	± 200mV to ± 2V	± 200mV to ± 2V	± 2V	± 200mV to ± 2V	± 200mV to ± 2V	± 200mV to ± 2V
Impedance	10 <sup>12</sup> Ω	10 <sup>12</sup> Ω	10 <sup>12</sup> Ω	10 <sup>12</sup> Ω	10 <sup>12</sup> Ω	10 <sup>12</sup> Ω
Leakage Current	2pA	2pA	3pA	1pA	2pA	3pA
Noise (peak-to-peak)	15µV typ.	15µV typ.	15µV typ.	7µV typ.	15µV typ.	15µV typ.
<b>Digital Input</b>	—	—	—	Decimal Points Continuity Hold, Range Select	Display Hold (7116)	Display Hold (7117)
<b>Digital Outputs</b>						
Format	Direct 7 Segment LCD Display	Direct 7 Segment LCD Display	Multiplex BCD	4½ Digit Triplexed LCD Display Drive w/Decimal Points, Low Battery and Continuity Indicators	Direct 7 Segment LCD Display AC: 4.5V Down from V+	Direct 7 Segment LED Display Comm Anode DTL/TTL/CMOS
Logic Level	AC: 4.5V Down from V+	AC: 4.5V Down from V+	TTL/CMOS			
<b>Power Supply</b>						
Voltage	+9V	± 5V	± 5V	+9V	+9V	± 5V
Current	100µA	200µA	1.8mA	1.8mA	1.8mA	1.8mA
Package	40 pin DIP	40 pin DIP	28 pin DIP	40 pin DIP	40 pin DIP	40 pin DIP

\*Also available LD110/111/114 (not recommended for new designs), and ICL7126 (recommended use ICL7136).

**Integrating Analog-to-Digital Converters for Data Acquisition**

Type	Single Chip	Two Chip System***	
	ICL7109	ICL8052A/8068 ICL7104-14	ICL8052A/8068 ICL7104-16
<b>Resolution</b>	±12-Bit Binary	±14-Bit	±16-Bit
<b>Accuracy</b>	±1 Count	±1 Count	±1 Count
<b>Microprocessor Compatible</b>	Yes	Yes	Yes
<b>Output</b>	Programmable: 1. Latched parallel 3 state Binary  2. Controlled 2-8 bit bytes	Programmable: 1. Latched parallel 3 state Binary  2. Controlled 2-8 Bit Byte for ICL7104-12/14 3-8 Bit Byte for ICL7104-16	
<b>Control Lines</b>	Run/Hold, Busy, Byte Enables, Mode, Load, Send Enable, Out of Range		
<b>Conversion Time</b>	10ms	80ms	330ms
<b>UART Compatible</b>	Yes	Yes	Yes
<b>Noise (Typical)</b>	15µV	2µV (8068)	2µV (8068)
<b>Input Current</b>	10pA	30pA (8052)	30pA (8052)
<b>Input Voltage Range</b>	+400mV to +4.1V	+100mV to +10V	-200mV to -10V

\*\*\*ICL8052/8068 and ICL8053 can be combined as analog portion of dual-slope A/D converter under µP control. See ICL8052/8068 and ICL 7104-16 for performance characteristics.

### Digital-to-Analog Converters\*

Maximum Electrical Specification at 25°C unless otherwise noted.

Model	ICL7134U/B	7145	7146	AD7523	AD7533	AD7520 (7530)	AD7520 (7531)	AD7541
<b>Resolution</b>	14 bit	16 bit	12 bit	8 bit	10 bit	10 bit	12 bit	12 bit
<b>Accuracy</b>	J/K/L	J/K	J/K	J/K/L	J/K/L	J/K/L	J/K/L	J/K/L
Linearity	0.01/0.006/0.003%	0.006/0.003%	0.01%	0.2%/0.1%/0.05%	0.2%/0.1%/0.05%	0.02%/0.01%/0.05%	0.2%/0.1%/0.5%	0.02%/0.01%/0.01%
Zero Offset	10 nA	10 mV	120 μV	50 μA	200 nA	200 nA (300 nA)	200 nA (300 nA)	50 nA
Full Scale Reading	0.003%	0.04/0.02% FSR	0.04/0.02% FSR	1.5% max	1.4%	0.3% typ	0.3% typ	0.3%
<b>Stability</b>								
Gain vs. Temperature	5 ppm/°C	1 ppm/°C typ	5 ppm/°C typ	10 ppm/°C	10 ppm/°C	10 ppm/°C	10 ppm/°C	10 ppm/°C
Linearity vs. Temperature	1 ppm/°C	1 ppm/°C typ	1 ppm/°C typ	2 ppm/°C	2 ppm/°C	2 ppm/°C	2 ppm/°C	2 ppm/°C
<b>Setting Time</b>								
To 1/2 LSB	0.9 μs typ	3 μs	10 μs	150 ns	600 ns typ	500 ns typ	500 ns typ	1 μs
<b>Input Code</b>	DTL/TTL/CMOS	DTL/TTL/CMOS	DTL/TTL/CMOS	DTL/TTL/CMOS	DTL/TTL/CMOS	DTL/TTL/CMOS	DTL/TTL/CMOS	DTL/TTL/CMOS
Logic Compatibility	Binary (U)	Binary or 2's Complement	Binary or 2's Complement	Binary or 2's Complement	Binary or 2's Complement	Binary or 2's Complement	Binary or 2's Complement	Binary or 2's Complement
option	2's Complement (B)	2's Complement	2's Complement	Offset Binary	Offset Binary	Offset Binary	Offset Binary	Offset Binary
<b>Power Supply</b>								
Voltage	+3.5 to +6.0V	4.5 to 5.5V	±4.5 to 5.5V	+6 to +16V	+5 to +15V	+5 to +15V	+5 to +15V	+5 to +16V
Current	2mA	1.2mA	5mA	100 μA	2mA	2mA	2mA	2mA
<b>Package</b>	28 pin DIP	28 pin DIP	28 pin DIP	16 pin DIP	16 pin DIP	16 pin DIP	18 pin DIP	18 pin DIP

\*R2R Ladder Multiplying Type

### Successive Approximation Analog-to-Digital Converters

Model	ADC0801-4	ICL7115
<b>Resolution</b>	8 bit	14 bit binary
<b>Accuracy</b>	± 1/4 / 1/2 / 1/2 / 1 LSB	± 1/2 LSB
<b>Microprocessor Compatible</b>	Yes	Yes
<b>Output</b>	Programmable: 1. Latched parallel 3 <sup>1</sup> state Binary 2. One 8 bit byte	Programmable: 1. Two latched bytes 2. 15 bit parallel
<b>Control Lines</b>	CS, RD, WR	CS, RD, WR, AO, BUS
<b>Conversion Time</b>	100 μs	40 μs
<b>UART Compatible</b>	Yes	No
<b>Input Voltage Range</b>	5V span	0 - 5V

### Quad Current Switches ICL8018/8019/8020

High speed precision current switches for use in current summing D/A converters. Can be purchased individually or in matched sets with accuracies of 0.01% (ICL8018), 0.1% (ICL8019), or 1.0% (ICL8020)

### Sample and Hold

Type	V <sub>analog</sub> (V <sub>p-p</sub> )	I <sub>acc</sub> ** (μA)	V <sub>infect</sub> ** (mV)	V <sub>os</sub> (mV)	Drift Rate (mV/sec)
IHS110	±7.5	6	5	40	5
IHS111	±10	6	5	40	5
IHS112	±7.5	6	5	10	5
IHS113	±10	6	5	10	5
IHS114	±7.5	6	5	5	5
IHS115	±10	6	5	5	5

\*\*C<sub>STO</sub> = 0.01 μF

### Monolithic Voltage Converter—The ICL7660

Converts positive voltage into negative over a range of +1.5V through +10V. May be cascaded for higher negative output voltages, paralleled for greater output current, used as a positive voltage multiplier, or any combination of the above. Typical supply current is 170 μA, and output source resistance is 55Ω at T<sub>A</sub> = 25°C and I<sub>O</sub> = 20 mA

## ANALOG SWITCHES & MULTIPLEXERS



### Analog Switches with Driver

Type	No. of Channels	Intersil Device No.	Switch Technology	$t_{DS(on)}$ ns max(1)	$I_{D(Off)}$ mA max	$t_{on}$ $\mu$ s max	$t_{off}$ $\mu$ s max	Logic Input			Input Type(2)	Power Consumption mW		
								Logic Level						
SPST	1	IH5021	P-JFET	100	0.2	0.5	0.5	TTL	High Level		lo			
		IH5022	P-JFET	150	0.2	0.5	0.5	TTL	Low Level		lo			
		IH5023	P-JFET	100	0.2	0.5	0.5	TTL	High Level		lo			
		IH5024	P-JFET	150	0.2	0.5	0.5	TTL	Low Level		lo			
		IH5037	P-JFET	100	0.5	0.2	0.2	TTL	High Level		lo			
		IH5038	P-JFET	150	0.5	0.2	0.2	TTL	High Level		lo			
		IH5040	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	hi	.035
IH5140	CMOS	50	0.1	0.1	0.075	TTL	CMOS				hi	.035		
SPST	2	DG180	N-JFET	10	10.0	0.3	0.25	DTL	TTL	RTL		lo	120	
		DG181	N-JFET	30	1.0	0.15	0.13	DTL	TTL	RTL		lo	120	
		DG182	N-JFET	75	1.0	0.25	0.13	DTL	TTL	RTL		lo	120	
		DGM182	CMOS	75	0.1	0.25	0.13	DTL	TTL	RTL		lo	.035	
		DG200	CMOS	70	1.0	0.7	0.5	DTL	TTL	RTL	CMOS	TTL High Level	lo	3.0
		IH5017	P-JFET	100	0.2	0.5	0.5	TTL	High Level			lo		
		IH5018	P-JFET	150	0.2	0.5	0.5	TTL	Low Level			lo		
		IH5019	P-JFET	100	0.2	0.5	0.5	TTL	High Level			lo		
		IH5020	P-JFET	150	0.2	0.5	0.5	TTL	Low Level			lo		
		IH5033	P-JFET	100	0.5	0.2	0.2	TTL	High Level			lo		
		IH5034	P-JFET	150	0.5	0.2	0.2	TTL	High Level			lo		
		IH5035	P-JFET	100	0.5	0.2	0.2	TTL	High Level			lo		
		IH5036	P-JFET	150	0.5	0.2	0.2	TTL	High Level			lo		
		IH5041	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	hi	.035
IH5048	CMOS	35	1.0	0.25	0.15	DTL	TTL	RTL	CMOS	PMOS	hi	.035		
IH5141	CMOS	50	0.1	0.1	0.075	TTL	CMOS				hi	.035		
IH5341	CMOS	75	0.1	0.3	0.15	TTL	CMOS				hi	.035		
SPST	3	IH5013	P-JFET	100	0.2	0.5	0.5	TTL	High Level		lo			
		IH5014	P-JFET	150	0.2	0.5	0.5	TTL	Low Level		lo			
		IH5015	P-JFET	100	0.2	0.5	0.5	TTL	High Level		lo			
		IH5016	P-JFET	150	0.2	0.5	0.5	TTL	Low Level		lo			
		IH5029	P-JFET	100	0.5	0.2	0.2	TTL	High Level		lo			
		IH5030	P-JFET	150	0.5	0.2	0.2	TTL	High Level		lo			
		IH5031	P-JFET	100	0.5	0.2	0.2	TTL	High Level		lo			
IH5032	P-JFET	150	0.5	0.2	0.2	TTL	High Level		lo					
SPST	4	DC118	P-MOSFET	450	-4.0	0.3	1.0	DTL	TTL	RTL		lo	133	
		DG201	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS		lo	.350
		IH5009	P-JFET	100	0.2	0.5	0.5	TTL	High Level			lo		
		IH5010	P-JFET	150	0.2	0.5	0.5	TTL	Low Level			lo		
		IH5011	P-JFET	100	0.2	0.5	0.5	TTL	High Level			lo		
		IH5011	P-JFET	150	0.2	0.5	0.5	TTL	Low Level			lo		
		IH5025	P-JFET	100	0.5	0.2	0.2	TTL	High Level			lo		
		IH5026	P-JFET	150	0.5	0.2	0.2	TTL	High Level			lo		
		IH5027	P-JFET	100	0.5	0.2	0.2	TTL	High Level			lo		
		IH5028	P-JFET	150	0.5	0.2	0.2	TTL	High Level			lo		
IH5052	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	hi	.350		
IH5053	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	hi	.350		
SPST	5	DG123	P-MOSFET	450	-4.0	0.3	1.0	DTL	TTL	RTL		hi	133	
		DG125	P-MOSFET	450	-4.0	0.3	1.0	DTL	TTL	RTL		lo	133	
		DG143A	N-JFET	80	1.0	0.4	0.8	DTL	TTL	RTL		(3)	84	
		DG144A	N-JFET	30	1.0	0.4	0.8	DTL	TTL	RTL		(3)	84	
		DG146A	N-JFET	10	10.0	0.5	1.25	DTL	TTL	RTL		(3)	84	
		DG161A	N-JFET	15	10.0	0.5	1.25	DTL	TTL	RTL		(3)	90	
		DG162A	N-JFET	50	2.0	0.4	0.8	DTL	TTL	RTL		(3)	90	
		DG186	N-JFET	10	10.0	0.3	0.25	DTL	TTL	RTL		(3)	73	
DG187	N-JFET	30	1.0	0.15	0.13	DTL	TTL	RTL		(3)	73			
SPDT	1	DG188	N-JFET	75	1.0	0.25	0.13	DTL	TTL	RTL		(3)	73	
		DGM188	CMOS	75	0.1	0.25	0.13	DTL	TTL	RTL		(3)	.035	
		IH5042	CMOS	75	1.0	0.05	0.025	DTL	TTL	RTL	PMOS	CMOS	(3)	.035
		IH5050	CMOS	35	1.0	0.25	0.15	DTL	TTL	RTL	PMOS	CMOS	(3)	.035
IH5142	CMOS	50	0.1	0.175	0.125	TTL	CMOS			(3)	.035			

### Analog Switches with Driver continued

Type	No. of Channels	Device No.	Switch Technology	$r_{DS(on)}$ $\Omega$ max(1)	$I_{D(off)}$ nA max	$t_{on}$ $\mu s$ max	$t_{off}$ $\mu s$ max	Logic Input			Input Type(2)	Power Consumption mW		
								Logic Level						
SPDT	2	DG189	N-JFET	10	10.0	0.3	0.25	DTL	TTL	RTL	(3)	120		
		DG190	N-JFET	30	1.0	0.15	0.13	DTL	TTL	RTL	(3)	120		
		DG191	N-JFET	75	1.0	0.25	0.13	DTL	TTL	RTL	(3)	120		
		DGM191	CMOS	75	0.1	0.25	0.13	DTL	TTL	RTL	(3)	.035		
		IH5043	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	PMOS	CMOS	(3)	.035
		IH5051	CMOS	35	1.0	0.25	0.15	DTL	TTL	RTL	PMOS	CMOS	(3)	.035
DPST	1	IH5044	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	hi	.035
		IH5144	CMOS	50	0.1	0.175	0.125	TTL	CMOS				hi	.035
DPST	2	DG183	N-JFET	10	10.0	0.3	0.25	DTL	TTL	RTL			hi	84
		DG184	N-JFET	30	1.0	0.15	0.13	DTL	TTL	RTL			hi	84
		DG185	N-JFET	75	1.0	0.25	0.13	DTL	TTL	RTL			hi	84
		DGM185	CMOS	75	0.1	0.25	0.13	DTL	TTL	RTL			hi	.035
		IH5045	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	PMOS	CMOS	hi	.035
		IH5049	CMOS	35	1.0	0.25	0.15	DTL	TTL	RTL	PMOS	CMOS	hi	.035
DPDT	1	IH5145	CMOS	50	0.1	0.175	0.125	TTL	CMOS				hi	.035
		DG139A	N-JFET	30	1.0	0.4	0.8	DTL	TTL	RTL			(3)	84
		DG142A	N-JFET	80	1.0	0.4	0.8	DTL	TTL	RTL			(3)	84
		DG145A	N-JFET	10	10.0	0.5	1.25	DTL	TTL	RTL			(3)	84
		DG163A	N-JFET	15	10.0	0.5	1.25	DTL	TTL	RTL			(3)	90
		DG164A	N-JFET	50	2.0	0.4	0.8	DTL	TTL	RTL			(3)	90
4PST	1	IH5046	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	(3)	.035
		IH5047	CMOS	75	1.0	0.5	0.25	DTL	TTL	RTL	CMOS	PMOS	hi	.035

### Multiplexers

Type	No. of Channels	Device No.	Switch Technology	$r_{DS(on)}$ $\Omega$ max(1)	$I_{D(off)}$ nA max	$t_{on}$ $\mu s$ max	$t_{off}$ $\mu s$ max	Logic Input			Input Type(2)	Power Consumption mW	
								Logic Level					
CMOS	1 of 8	IH6108	CMOS	300	0.1	1.5	1.0	DTL	TTL	RTL	CMOS	hi	4.5
CMOS	1 of 16	IH6116	CMOS	600	0.2	1.5	1.0	DTL	TTL	RTL	CMOS	hi	4.5
CMOS	2 of 8	IH6208	CMOS	300	0.1	1.5	1.0	DTL	TTL	RTL	CMOS	hi	4.5
CMOS	2 of 16	IH6216	CMOS	600	0.2	1.5	1.0	DTL	TTL	RTL	CMOS	hi	4.5
CMOS Fault Protected	1 of 8	IH5108	CMOS	700	0.1	1.5	1.0	DTL	TTL	RTL	CMOS	hi	4.5
	2 of 8	IH5208	CMOS	700	0.1	1.5	1.0	DTL	TTL	RTL	CMOS	hi	4.5

### Drivers for FET Switches

Electrical Characteristics @ +25°C—Military Temperature Devices

No. of Channels	Device No.	$V_{OUT}$		$t_{on}$ ns max	$t_{off}$ ns max	$I_{INL}$ $\mu A$ (max)	$I_{INH}$ mA (max)	Logic Input Level	Power Consumption (mW)
		Positive Volts	Negative Volts						
2	IH6201	+14.0	-14.0	200	300	1.0	1.0	TTL	350
4	D129	$V_{supply}$	-19.3	250	1000	200	0.25 $\mu A$	TTL/DTL	55
6	D123	$V_{supply}$	-19.7	250	600	1.0	1.0 $\mu A$	TTL/DTL	20
	D125	$V_{supply}$	-19.7	250	600	1.0	1.5 $\mu A$	TTL	50

### RF/VIDEO SWITCH

Type	No. of Channels	Device No.	$r_{DS(ON)}$ $\Omega$ Max.	Off Isolation	Logic Input	Power Consumption (mW)
CMOS	2	IH5341	75	> 60dB @ 10 MHz	TTL, CMOS	0.030

**Notes:**

1. Switch Resistance under worst case analog voltage.
2. Positive logic LO ("0") or HI ("1") voltage at driver input necessary to turn switch on.
3. Logic "0" or "1" can be arbitrarily assigned for double-throw switches.
4. Switch resistance under best case analog voltage.





## Counters, Timers and Display Drivers

Part Number	Circuit Description	Package	Crystal Frequency	Output
ICM7207 ICM7207A	Frequency counter timebase.	14-Pin DIP 14-Pin DIP	6.5536 MHz 5.2488 MHz	0.01, 0.1, or 1-second count window plus store, reset and MUX.
ICM7208	7-digit unit counter. With addition of 7207 the circuit becomes a complete timer-frequency counter.	28-Pin DIP	—	LED display drive
ICM7211 ICM7212	Four-digit display decoder drivers: ICM7211 is LCD, ICM7212 is LED. Non-multiplexed for low noise. BCD input, decoded display drive output.	40-Pin DIP (plastic)	—	Four-digit, seven-segment direct display drive: LED or LCD
ICM7216 ICM7226	Eight-digit universal counter measures frequency, period, frequency ratio, time interval, units; on-board time base.	28-Pin DIP 40-Pin DIP (Cerdip or plastic)	1 or 10 MHz	Eight-digit-common anode or common cathode direct LED drive: BCD output
ICM7217 ICM7227	Four-digit CMOS up/down counter; presettable start/count and compare register; for hard-wired or microprocessor control applications; cascable.	28-Pin Cerdip or plastic	—	Four-digit, seven-segment common anode or common cathode direct LED display drive: equal, zero, carry/borrow
ICM7218A/D ICM7218E	LED display driver system with 8x8 memory; numeric or dot (1 of 64) decoding; microprocessor compatible.	28-Pin DIP 40-Pin DIP (Cerdip or plastic)	—	Eight-digit, seven-segment plus decimal point; common cathode or common anode
ICM7224 ICM7225	4½-digit high speed counter/decoder/driver. 25 MHz typ; ICM7224 is LCD, ICM7225 is LED; direct display drive, cascable.	40-Pin DIP (plastic)	—	4½-digit seven-segment direct display driver: LED or LCD
ICM7231	8-digit CMOS multiplexed LCD driver. Parallel input.	40-Pin DIP (plastic)	—	Eight-digit, seven-segment plus two flags per digit
ICM7232	10½-digit CMOS multiplexed LCD driver. Serial input.	40-Pin DIP (plastic)	—	10½-digit, seven-segment plus two flags per digit
ICM7233	4-character CMOS multiplexed LCD driver. Parallel alphanumeric (6-bit ASCII) input.	40-Pin DIP (plastic)	—	Four-character, 16-segment plus colon
ICM7234	5-character CMOS multiplexed LCD driver. Serial alphanumeric (6-bit ASCII) input.	40-Pin DIP (plastic)	—	Five-character, 16-segment plus colon
ICM7235/A	4-digit CMOS decoder/driver for direct drive vacuum fluorescent displays. BCD input.	40-Pin DIP (plastic)	—	Four-digit, seven-segment, vacuum fluorescent display drive: either HEX or CODE B
ICM7235M/AM	Same as above but microprocessor compatible.	—	—	—
ICM7236	4½-digit high speed CMOS counter/decoder/driver for vacuum fluorescent displays: 25 MHz typ. counting speed.	40-Pin DIP (plastic)	—	4½-digit, seven-segment, vacuum fluorescent display drive
ICM7236A	Same as above but counting to 15959.	40-Pin DIP (plastic)	—	4½-digit, seven-segment, vacuum fluorescent display drive
ICM7240 ICM7250 ICM7260	Programmable CMOS counter/timers using external RC time base. Programmable from $\mu$ s to years.	16-Pin DIP	External	Timed output
ICM7242	Fixed CMOS counter/timer. Uses external RC time base; sequence timing from $\mu$ s to minutes.	8-Pin DIP	External	Timed output
ICM7243	8-character multiplexed LED display driver with alphanumeric (6-bit ASCII) input.	40-Pin Cerdip	—	Eight-character, 14/16-segment common cathode alphanumeric LED display drive
ICM7281	LCD Dot Matrix Column Driver	40-Pin DIP	—	Up to 256 x 256 dots
ICM7555 ICM7556	Single or dual CMOS version of industry-standard 555 timer; 80 $\mu$ A typ. supply current; 500 kHz guaranteed; 2-18V power supply.	8-Pin DIP 14-Pin DIP	—	—



# ADC0801 – ADC0804

## 8-Bit Microprocessor Compatible A/D Converters

### FEATURES

- MCS-48 and MCS-80/85 bus compatible—no interfacing logic required
- Conversion time < 100 $\mu$ s
- Easy interface to all microprocessors
- Will operate “stand alone”
- Differential analog voltage inputs
- Works with bandgap voltage references
- TTL compatible inputs and outputs
- On-chip clock generator
- 0V to 5V analog voltage input range (single +5V supply)
- No zero-adjust required

### GENERAL DESCRIPTION

The ADC0801 family are CMOS 8-bit successive approximation A/D converters which use a modified potentiometric ladder, and are designed to operate with the 8080A control bus via three-state outputs. These converters appear to the processor as memory locations or I/O ports, hence no interfacing logic is required.

The differential analog voltage input has good common-mode-rejection, and permits offsetting the analog zero-input-voltage value. In addition, the voltage reference input can be adjusted to allow encoding any smaller analog voltage span to the full 8 bits of resolution.

The ADC0801 family is available in the industry standard 20 pin CERDIP package.

#### TYPICAL APPLICATION

8-BIT RESOLUTION  
OVER ANY DESIRED  
ANALOG INPUT  
VOLTAGE RANGE

#### PIN CONFIGURATION

TOP VIEW  
(Outline dwg. JP)

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#### ORDERING INFORMATION

PART	ERROR	TEMPERATURE RANGE	PACKAGE	ORDER NUMBER
ADC0801	$\pm 1/4$ bit adjusted full-scale	0°C to +70°C	20 pin CERDIP	ADC0801LCN
		-40°C to +85°C	20 pin CERDIP	ADC0801LCD
		-55°C to +125°C	20 pin CERDIP	ADC0801LD
ADC0802	$\pm 1/2$ bit no adjust	0°C to +70°C	20 pin CERDIP	ADC0802LCN
		-40°C to +85°C	20 pin CERDIP	ADC0802LCD
		-55°C to +125°C	20 pin CERDIP	ADC0802LD
ADC0803	$\pm 1/2$ bit adjusted full-scale	0°C to +70°C	20 pin CERDIP	ADC0803LCN
		-40°C to +85°C	20 pin CERDIP	ADC0803LCD
		-55°C to +125°C	20 pin CERDIP	ADC0803LD
ADC0804	$\pm 1$ bit no adjust	0°C to +70°C	20 pin CERDIP	ADC0804LCN
		-40°C to +85°C	20 pin CERDIP	ADC0804LCD

## ADC0801-ADC0804



### ABSOLUTE MAXIMUM RATINGS

Supply Voltage	6.5V
Voltage at Any Input	-0.3V to (V <sup>+</sup> + 0.3V)
Storage Temperature Range	-65°C to +150°C
Package Dissipation at T <sub>A</sub> = +25°C	875 mW
Lead Temperature (Soldering, 10 seconds)	300°C

### OPERATING RATINGS

Temperature Range	
ADC0801/02/03LD	-55°C to +125°C
ADC0801/02/03/04LCD	-40°C to +85°C
ADC0801/02/03/04LCN	0°C to +70°C
Supply Voltage Range	4.5V to 6.3V

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### SYSTEM ELECTRICAL CHARACTERISTICS (Notes 1 and 7)

Converter Specifications: V<sup>+</sup> = 5V, V<sub>REF/2</sub> = 2.500V, T<sub>MIN</sub> ≤ T<sub>A</sub> ≤ T<sub>MAX</sub> and f<sub>CLK</sub> = 640kHz unless otherwise stated.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ADC0801: Total Adjusted Error	With Full Scale Adjust			±1/4	LSB
ADC0802: Total Unadjusted Error	Completely Unadjusted			±1/2	LSB
ADC0803: Total Adjusted Error	With Full Scale Adjust			±1/2	LSB
ADC0804: Total Unadjusted Error	Completely Unadjusted			±1	LSB
V <sub>REF/2</sub> Input Resistance	Input Resistance at Pin 9	1.0	1.3		kΩ
Analog Input Voltage Range	(Note 2)	GND - 0.05		V <sup>+</sup> + 0.05	V
DC Common-Mode Rejection	Over Analog Input Voltage Range		±1/16	±1/8	LSB
Power Supply Sensitivity	V <sup>+</sup> = 5V ±10% Over Allowed Input Voltage Range		±1/16	±1/8	LSB

### AC ELECTRICAL CHARACTERISTICS

Timing Specifications: V<sup>+</sup> = 5V and T<sub>A</sub> = +25°C unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Clock Frequency	f <sub>CLK</sub>	V <sup>+</sup> = 6V (Note 3)	100	640	1280	kHz
		V <sup>+</sup> = 5V	100	640	800	kHz
Clock Periods per Conversion (Note 4)	t <sub>conv</sub>		66		73	
Conversion Rate In Free-Running Mode	CR	.INTR tied to WR with CS = 0V, f <sub>CLK</sub> = 640kHz			8888	conv/s
Width of WR Input (Start Pulse Width)	t <sub>W(WR)I</sub>	CS = 0V (Note 5)	100			ns
Access Time (Delay from Falling Edge of RD to Output Data Valid)	t <sub>acc</sub>	C <sub>L</sub> = 100pF (Use Bus Driver IC for Larger C <sub>L</sub> )		135	200	ns
3-State Control (Delay from Rising Edge of RD to Hi-Z State)	t <sub>th</sub> , t <sub>oh</sub>	C <sub>L</sub> = 10pF, R <sub>L</sub> = 10k (See 3-State Test Circuits)		125	250	ns
Delay from Falling Edge of WR to Reset of INTR	t <sub>WI</sub> , t <sub>RI</sub>			300	450	ns
Input Capacitance of Logic Control Inputs	C <sub>IN</sub>			5	7.5	pF
3-State Output Capacitance (Data Buffers)	C <sub>OUT</sub>			5	7.5	pF



# ICL7106/7107

## 3½-Digit Single Chip A/D Converter

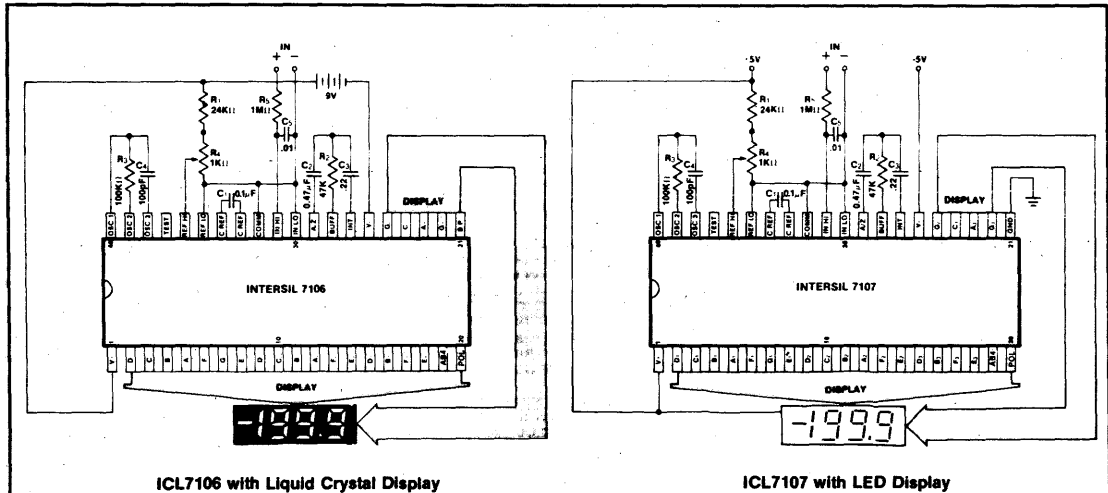
### FEATURES

- **Guaranteed zero reading for 0 volts input on all scales.**
- **True polarity at zero for precise null detection.**
- **1 pA typical input current.**
- **True differential input and reference.**
- **Direct display drive - no external components required. — LCD ICL7106  
— LED ICL7107**
- **Low noise - less than 15µV p-p.**
- **On-chip clock and reference.**
- **Low power dissipation - typically less than 10mW.**
- **No additional active circuits required.**
- **Evaluation Kit available.**

### GENERAL DESCRIPTION

The Intersil ICL7106 and 7107 are high performance, low power 3½-digit A/D converters containing all the necessary active devices on a single CMOS I.C. Included are seven-segment decoders, display drivers, reference, and a clock. The 7106 is designed to interface with a liquid crystal display (LCD) and includes a backplane drive; the 7107 will directly drive an instrument-size light emitting diode (LED) display.

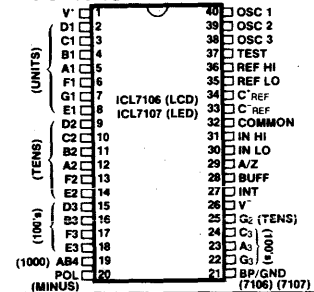
The 7106 and 7107 bring together an unprecedented combination of high accuracy, versatility, and true economy. High accuracy like auto-zero to less than 10µV, zero drift of less than 1µV/°C, input bias current of 10 pA max., and roll-over error of less than one count. The versatility of true differential input and reference is useful in all systems, but gives the designer an uncommon advantage when measuring load cells, strain gauges and other bridge-type transducers. And finally the true economy of single power supply operation (7106), enabling a high performance panel meter to be built with the addition of only 7 passive components and a display.



### ORDERING INFORMATION

Part	Package	Temp. Range	Order Part #
7106	40 pin ceramic DIP	0°C to +70°C	ICL7106CDL
7106	40 pin plastic DIP	0°C to +70°C	ICL7106CPL
7106	40 pin Cerdip	0°C to +70°C	ICL7106CJL
7107	40 pin Cerdip	0°C to +70°C	ICL7107CJL
7107	40 pin ceramic DIP	0°C to +70°C	ICL7107CDL
7107	40 pin plastic DIP	0°C to +70°C	ICL7107CPL
7106 Kit	Evaluation kits contain IC, display, circuit board, passive components and hardware.		ICL7106EV/Kit
7107 Kit			ICL7107EV/Kit

### PIN CONFIGURATION





# ICL7109 12 Bit Binary A/D Converter for Microprocessor Interfaces

## FEATURES

- 12 bit binary (plus polarity and overrange) dual slope integrating analog-to-digital converter.
- Byte-organized TTL-compatible three-state outputs and UART handshake mode for simple parallel or serial interfacing to microprocessor systems.
- RUN/HOLD input and STATUS output can be used to monitor and control conversion timing.
- True differential input and differential reference.
- Low noise — typically 15 $\mu$ V p-p.
- 1pA typical input current.
- Operates at up to 30 conversions per second.
- On-chip oscillator operates with inexpensive 3.58MHz TV crystal giving 7.5 conversions per second for 60Hz rejection. May also be operated as RC oscillator for other clock frequencies.
- Fabricated using MAX-CMOS™ technology combining analog and digital functions on a single low power LSI CMOS chip.
- All inputs fully protected against static discharge; no special handling precautions necessary.

## GENERAL DESCRIPTION

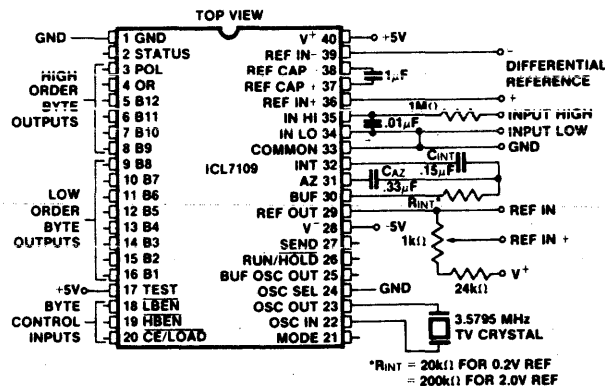
The ICL7109 is a high performance, low power integrating A/D converter designed to easily interface with microprocessors.

The output data (12 bits, polarity and overrange) may be directly accessed under control of two byte enable inputs and a chip select input for a simple parallel bus interface. A UART handshake mode is provided to allow the ICL7109 to work with industry-standard UARTs in providing serial data transmission, ideal for remote data logging applications. The RUN/HOLD input and STATUS output allow monitoring and control of conversion timing.

The ICL7109 provides the user with the high accuracy, low noise, low drift, versatility and economy of the dual-slope integrating A/D converter. Features like true differential input and reference, drift of less than 1 $\mu$ V/°C, maximum input bias current of 10pA, and typical power consumption of 20mW make the ICL7109 an attractive per-channel alternative to analog multiplexing for many data acquisition applications.

## PIN CONFIGURATION AND TEST CIRCUIT:

(See Figure 1 for typical connection to a UART or Microcomputer)



(OUTLINE DWGS DL, JL, PL)

## ORDERING INFORMATION

Part	Temp. Range	Package	Order Number
7109	-55°C to +125°C	40-Pin Ceramic DIP	ICL7109MDL
7109	-20°C to +85°C	40-Pin Ceramic DIP	ICL7109IDL
7109	-20°C to +85°C	40-Pin CERDIP	ICL7109JL
7109	0°C to 70°C	40-Pin Plastic DIP	ICL7109CPL



# ICL7115

## Fast CMOS Monolithic 14-Bit A/D Converter

**PRELIMINARY**  
Specifications Subject To Change Without Notice

Intersil  
INTERFACE

### FEATURES

- 14-bit linearity and resolution (0.003%)
- No missing codes
- Microprocessor compatible byte-organized buffered outputs
- Fast conversion (40 $\mu$ s)
- Auto-zeroed comparator for low offset voltage
- Low linearity and gain tempco (1ppm/ $^{\circ}$ C, 4ppm/ $^{\circ}$ C)
- Low power consumption (60mW)
- No gain or offset adjustment necessary (0.006% FS)
- Provides 3% useable overrange
- FORCE/SENSE and separate digital and analog ground pins for increased system accuracy

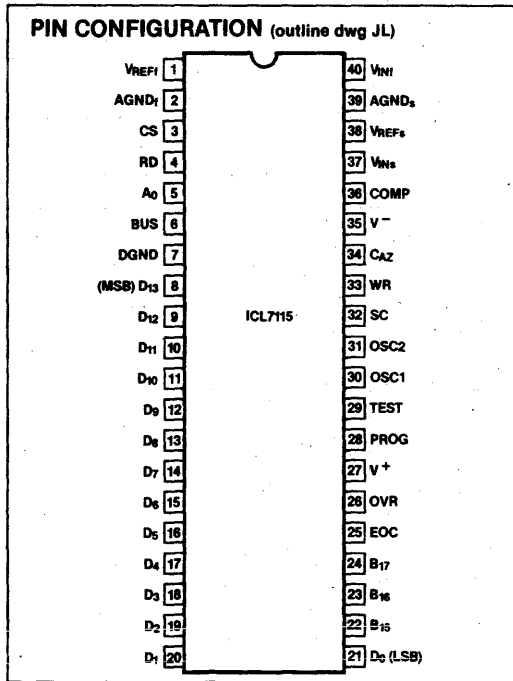
### GENERAL DESCRIPTION

The ICL7115 is the first monolithic 14-bit accurate, fast successive approximation A/D converter. It uses thin film resistors and CMOS circuitry with an on-chip PROM calibration table circuit to achieve 14-bit linearity without laser trimming. Special design techniques used in the DAC and comparator result in high speed, while the fully static silicon-gate CMOS circuitry keeps the power dissipation very low.

Microprocessor bus interfacing is made easy by the use of standard WRite and ReaD cycle timing and control signals, combined with Chip Select and Address pins. The digital output pins are byte-organized and three-state gated for bus interface to 8, 12, and 16-bit systems.

The ICL7115 provides separate Analog and Digital grounds. Analog ground, voltage reference and input voltage pins are separated into force and sense lines for increased system accuracy. Operating with  $\pm 5$ V supplies, the ICL7115 accepts 0V to +5V input with a -5V reference or 0V to -5V input with a +5V reference.

The ICL7115 is available in several versions with different accuracies, temperature ranges and packages. A Leadless Chip Carrier (LCC) package is also available; consult factory.



### ORDERING INFORMATION

ACCURACY	PACKAGE	TEMPERATURE	PART NUMBER
0.01%	40-Pin Cerdip	0 $^{\circ}$ C to +70 $^{\circ}$ C	ICL7115JCJL
0.01%	40-Pin Cerdip	-25 $^{\circ}$ C to +85 $^{\circ}$ C	ICL7115JIJL
0.01%	LCC	-	-
0.006%	40-Pin Cerdip	0 $^{\circ}$ C to +70 $^{\circ}$ C	ICL7115KCJL
0.006%	40-Pin Cerdip	-25 $^{\circ}$ C to +85 $^{\circ}$ C	ICL7115KIJL
0.006%	LCC	-	-
0.003%	40-Pin Cerdip	0 $^{\circ}$ C to +70 $^{\circ}$ C	ICL7115LCJL
0.003%	40-Pin Cerdip	-25 $^{\circ}$ C to +85 $^{\circ}$ C	ICL7115LIJL
0.003%	LCC	-	-

**ICL7115**



**ABSOLUTE MAXIMUM RATINGS** (Note 1)

Supply Voltage  $V^+$  to DGND..... -0.3V to +6.5V  
 Supply Voltage  $V^-$  to DGND..... +0.3V to -6.5V  
 $V_{REFS}$ ,  $V_{REFI}$ ,  $V_{INS}$ ,  $V_{INF}$  to DGND..... +25V to -25V  
 $AGND_S$ ,  $AGND_I$  to DGND..... +1V to -1V  
 Current in FORCE and SENSE Lines..... 25mA  
 Digital I/O Pin Voltages..... -0.3V to  $V^+$  +0.3V  
 PROG to DGND Voltage.....  $V^-$  to  $V^+$  +0.3V

Operating Temperature Range  
 ICL7115XCXX..... 0°C to +70°C  
 ICL7115XIXX..... -25°C to +85°C  
 Storage Temperature Range..... -65°C to +150°C  
 Power Dissipation..... 500mW  
 derate above 70°C @ 100mW/°C  
 Lead Temperature (Soldering, 10 sec)..... 300°C  
**Note 1:** All voltages with respect to DGND, unless otherwise noted.

*Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.*

**DC ELECTRICAL CHARACTERISTICS**  $V^+ = +5.0V$ ,  $V^- = -5.0V$ ,  $V_{REFS} = +5.0V$ ,  $T_A = +25^\circ C$  unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNIT
Resolution		$\overline{SC} = \text{High}$ $\overline{SC} = \text{Low}$	14 12			Bits
Total Unadjusted Error					1	LSB
Differential Non-Linearity		Full Operating Temperature Range		1/2		
Overall Accuracy (Note 3)		ICL7115J ICL7115K ICL7115L			0.01 0.006 0.003	% FSR
Full-Scale Error		$T_A = +25^\circ C$ Operating Temperature Range (Note 2)		1/2 1	1 4	LSB ppm/°C
Zero Error		$T_A = +25^\circ C$ Operating Temperature Range (Note 2)			1/8 4	LSB ppm/°C
Power Supply Rejection	PSRR	$T_A = +25^\circ C$ Full Operating Temperature Range		1/16	1/8	LSB
$V_{INS}$ , $V_{REFS}$ Resistance	$Z_{IN}$ , $Z_{REF}$	(Note 4) Operating Temperature Range	3	5 -300	7	k $\Omega$ ppm/°C
Low State Input Voltage	$V_{il}$	Operating Temperature Range			0.8	V
High State Input Voltage	$V_{ih}$	Operating Temperature Range	2.4			V
Logic Input Current	$I_{ih}$	$0 < V_{IN} < V^+$		1	10	$\mu A$
Low State Output Voltage	$V_{ol}$	$I_{OUT} = 3.2mA$ Operating Temperature Range			0.4	V
High State Output Voltage	$V_{oh}$	$I_{OUT} = -200\mu A$ Operating Temperature Range	2.8			V
Three-State Output Current	$I_{ox}$	$0 < V_{OUT} < V^+$		1		$\mu A$
Logic Input Capacitance	$C_{in}$	(Note 2)		15		pF
Logic Output Capacitance	$C_{out}$	Three-State (Note 2)		15		
Supply Voltage Range	$V^+$	Functional Operation	4.5		6.0	V
	$V^-$		-4.5		-6.0	
Supply Current	$I^+$	Excluding Ladder Current		5		mA
	$I^-$	$F_{CLK} = 1kHz$		5		

**Note 2:** Assumes all leads soldered or welded to printed circuit board.  
**Note 3:** Full-scale range (FSR) is 10V (+5V to -5V).  
**Note 4:** Guaranteed by design, not 100% tested in production.



**PRELIMINARY**  
Specifications Subject to Change Without Notice

# ICL7129

## 10-Digit Single-Chip A/D Converter

### FEATURES

- $\pm 19,999$  count A/D converter accurate to  $\pm 1$  count
- $10\mu\text{V}$  resolution on 200mV scale
- 110dB CMRR
- Direct LCD display drive
- True differential input and reference
- Low power consumption
- Decimal point drive outputs
- Overrange and underrange outputs
- Low battery detection and indication
- 10:1 range change input

### ORDERING INFORMATION

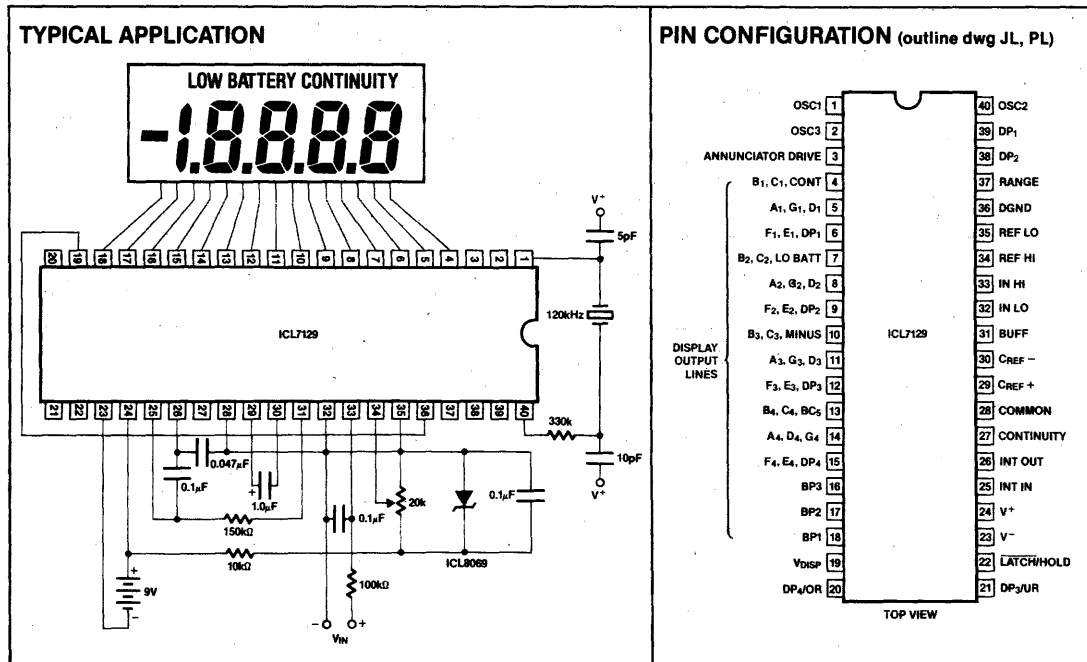
PART	PACKAGE	TEMPERATURE	ORDER NUMBER
7129	40-Pin CERDIP	0°C to +70°C	ICL7129CJL
7129	40-Pin Plastic	0°C to +70°C	ICL7129CPL
7129	40-Pin Plastic	0°C to +70°C	ICL7129RCPL
7129	Dice	0°C to +70°C	ICL7129C/D
7129	Flat Pack		

### GENERAL DESCRIPTION

The Intersil ICL7129 is a very high-performance  $4\frac{1}{2}$ -digit analog-to-digital converter that directly drives a multiplexed liquid crystal display. This single-chip CMOS integrated circuit requires only a few passive components and a reference to operate. And it is ideal for high-resolution hand-held digital multimeter applications.

The performance of the ICL7129 has not been equaled before in a single-chip A/D converter. The successive integration technique used in the ICL7129 results in accuracy better than 0.005% of full-scale and resolution down to  $10\mu\text{V}/\text{count}$ .

The ICL7129, drawing only 1mA from a 9V battery, is well suited for battery powered instruments. Provision has been made for the detection and indication of a "LOW BATTERY" condition. Autoranging instruments can be made with the ICL7129 which provides overrange and underrange outputs and 10:1 range changing input. The ICL7129 instantly checks for continuity, giving both a visual indication and a logic level output which can enable an external audible signal. These features and the high performance of the ICL7129 make it an extremely versatile and accurate instrument-on-a-chip.





# ICL7129



## ABSOLUTE MAXIMUM RATINGS

Supply Voltages ( $V^+$ to $V^-$ )	15V	Power Dissipation (Note 2)	
Reference Voltage (REF HI or REF LO)	$V^+$ to $V^-$	CERDIP package	1000mW
Input Voltage (Note 1) (IN HI or IN LO)	$V^+$ to $V^-$	Plastic package	800mW
$V_{DISP}$	$V^+$ to DGND - 0.3V	Operating Temperature	0°C to +70°C
Digital Input Pins		Storage Temperature	-65°C to +160°C
1, 2, 19, 20, 21, 22, 27,		Lead Soldering Temperature	300°C
37, 38, 39, 40	DGND to $V^+$		

**Note 1:** Input voltages may exceed the supply voltages provided that input current is limited to  $\pm 400\mu A$ . Currents above this value may result in invalid display readings but will not destroy the device if limited to  $\pm 1mA$ .

**Note 2:** Dissipation ratings assume device is mounted with all leads soldered to printed circuit board.

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS $V^+$ to $V^- = 9V$ , $V_{REF} = 1.00V$ , $T_A = +25^\circ C$ , $f_{CLK} = 120kHz$ , unless otherwise noted.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNIT
Zero Input Reading	$V_{IN} = 0V$ 200mV Scale	-0000	0000	+0000	Reading
Zero Reading Drift	$V_{IN} = 0V$ $0^\circ C < T_A < +70^\circ C$		$\pm 0.5$		$\mu V/^\circ C$
Ratiometric Reading	$V_{IN} = V_{REF} = 1000mV$ RANGE = 2V	9998	9999	10000	Reading
Range Change Accuracy	$V_{IN} = 0.10000V$ on Low Range + $V_{IN} = 1.0000V$ on High Range	0.9999	1.0000	1.0001	Ratio
Rollover Error	$-V_{IN} = +V_{IN} = 199mV$		0.5	1.0	Counts
Linearity Error	200mV Scale		0.5		
Input Common-Mode Rejection Ratio	$V_{CM} = 1.0V$ , $V_{IN} = 0V$ 200mV Scale		110		dB
Input Common-Mode Voltage Range	$V_{IN} = 0V$ 200mV Scale	$(V^-) + 1.5$		$(V^+) - 0.5$	V
Noise (p-p Value not Exceeding 95% of Time)	$V_{IN} = 0V$ 200mV Scale		7.0		$\mu V$
Input Leakage Current	$V_{IN} = 0V$ , Pin 32, 33		1	10	pA
Scale Factor Tempco	$V_{IN} = 199mV$ $0^\circ C < T_A < +70^\circ C$ External $V_{REF} = 0ppm/^\circ C$		2	5	ppm/°C
COMMON Voltage	$V^+$ to Pin 28	2.8	3.2	3.5	V
COMMON Sink Current	$\Delta Common = +0.1V$		0.6		mA
COMMON Source Current	$\Delta Common = -0.1V$		12		$\mu A$
DGND Voltage	$V^+$ to Pin 36 $V^+$ to $V^- = 9V$	4.5	5.3	5.8	V
DGND Sink Current	$\Delta DGND = +0.5V$		1.2		mA
Supply Voltage Range	$V^+$ to $V^-$	6	9	14	V
Supply Current Excluding COMMON Current	$V^+$ to $V^- = 9V$		1.0	1.4	mA
Clock Frequency			120	360	kHz
Display Multiplex Rate	$f_{CLK} = 120kHz$		100		Hz
$V_{DISP}$ Resistance	$V_{DISP}$ to $V^+$		50		k $\Omega$



# ICL7134

## 14-Bit $\mu$ P-Compatible Multiplying D/A Converter

INTERFACE  
Intersil

### FEATURES

- 14-bit linearity (0.003% FSR)
- No gain adjustment necessary
- Microprocessor-compatible with double buffered inputs
- Bipolar application requires no extra adjustments or external resistors
- Output current settling-time 3 $\mu$ s max (0.9 $\mu$ s typ)
- Low linearity and gain temperature coefficients
- Low power dissipation
- Full four-quadrant multiplication
- Full temperature range operation

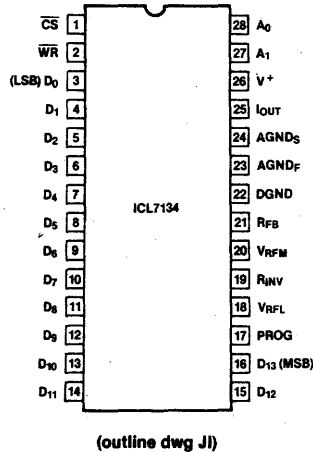
### GENERAL DESCRIPTION

The ICL7134 combines a four-quadrant multiplying DAC using thin film resistors and CMOS circuitry with an on-chip PROM-controlled correction circuit to achieve true 14-bit linearity without laser trimming.

Microprocessor bus interfacing is eased by standard memory WRite cycle timing and control signal use. Two input buffer registers are separately loaded with the 8 least significant bits (LS register) and the 6 most significant bits (MS register). Their contents are then transferred to the 14-bit DAC register, which controls the output switches. The DAC register can also be loaded directly from the data inputs, in which case the registers are transparent.

The ICL7134 is supplied in two versions. The ICL7134U is programmed for unipolar operation while the ICL7134B is programmed for bipolar applications. The  $V_{REF}$  input to the most significant bit of the DAC is separated from the reference input to the remainder of the ladder. For unipolar use, the two reference inputs are tied together, while for bipolar operation, the polarity of the MSB reference is reversed, giving the DAC a true 2's complement input transfer function. Two resistors which facilitate the reference inversion are included on the chip, so only an external op-amp is needed. The PROM is coded to correct for errors in these resistors as well as the inversion of the MSB.

### PIN CONFIGURATION



### ORDERING INFORMATION

NON-LINEARITY	TEMPERATURE RANGE		
	0°C to +70°C	-25°C to +85°C	-55°C to +125°C
Bipolar Versions			
0.01% (12-bit)	ICL7134BJCJI	ICL7134BJIJI	ICL7134BJMJI
0.006% (13-bit)	ICL7134BKCJI	ICL7134BKIJI	ICL7134BKMJI
0.003% (14-bit)	ICL7134BLCJI	ICL7134BLIJI	ICL7134BLMJI
Unipolar Versions			
0.01% (12-bit)	ICL7134UJCJI	ICL7134UJIJI	ICL7134UJMJI
0.006% (13-bit)	ICL7134UKCJI	ICL7134UKIJI	ICL7134UKMJI
0.003% (14-bit)	ICL7134ULCJI	ICL7134ULIJI	ICL7134ULMJI

Package: 28-pin Cerdip only

# ICL7134



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage ( $V^+$  to DGND) ..... -0.3V to 7.5V  
 $V_{RFL}$ ,  $V_{RFM}$ ,  $R_{INV}$ ,  $R_{FB}$  to DGND .....  $\pm 25V$   
 $I_{OUT}$ ,  $AGND_F$ ,  $AGND_S$  ..... -0.1V to  $V^+$   
 Current in  $AGND_S$ ,  $AGND_F$  ..... 25mA  
 $A_n$ ,  $D_n$ ,  $WR$ ,  $\overline{CS}$ ,  $PROG$  ..... -0.3V to  $V^+$  +0.3V  
 Operating Temperature Range  
 ICL7134XXC ..... 0°C to +70°C  
 ICL7134XXI ..... -20°C to +85°C  
 ICL7134XXM ..... -55°C to +125°C

Storage Temperature Range ..... -65°C to +150°C  
 Power Dissipation (Note 2) ..... 500mW  
 Derate Linearly Above 70°C @ 10mW/°C  
 Lead Temperature (Soldering, 10 seconds) ..... 300°C

**Note 1:** All voltages with respect to DGND.  
**Note 2:** Assumes all leads soldered or welded to printed circuit board.

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## OPERATING CHARACTERISTICS ( $V^+ = 5V$ , $V_{REF} = 10V$ , $T_A = +25^\circ C$ unless otherwise specified.)

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
Resolution			14			Bits
Non-Linearity	J	Test Figure 1 (Notes 1 and 2)			0.010	% FSR
	K				0.006	% FSR
	L				0.003	% FSR
Non-Linearity Temperature Coefficient		Operating Temperature Range		1	2	ppm/°C
Gain Error	J	Test Figure 1 (Notes 1 and 2)			0.020	% FSR
	K				0.012	% FSR
	L				0.006	% FSR
Gain Error Temperature Coefficient				2	8	ppm/°C
Monotonicity	J		14			Bits
	K		14			Bits
	L		14			Bits
$I_{OUT}$ Leakage Current	$I_{OLK}$	$T_A = +25^\circ C$			10	nA
		Operating Temperature Range		50		
Power Supply Rejection	PSRR	$T_A = +25^\circ C$ , $\Delta V^+ = \pm 10\%$		1	50	ppm/V
		Operating Temperature Range			100	
Output Current Settling Time				0.9	3	$\mu s$
Feedthrough Error	ICL7134U	$V_{REF} = \pm 10V$ , 2kHz Sinewave		250		$\mu V_{p-p}$
	ICL7134B			500		
Reference Input Resistance	$Z_{REF}$	$V_{RFL} = V_{RFM}$ (Unipolar Mode)	4.0		10	k $\Omega$
Output Capacitance	$C_{OUT}$	DAC Register = All 0's		160		pF
		DAC Register = All 1's		235		
Output Noise		Equivalent Johnson Res.		7		k $\Omega$
Low State Input	$V_{INL}$	Operating Temperature Range			0.8	V
High State Input	$V_{INH}$	Operating Temperature Range	2.4			V
Logic Input Current	$I_{in}$	$0 \leq V_{IN} \leq V^+$			1.0	$\mu A$
Logic Input Capacitance	$C_{in}$	(Note 3)		15		pF
Supply Voltage Range	$V^+$	Functional Operation	3.5		6.0	V
Supply Current	$I^+$	(Excluding Ladder)		0.06	0.5	mA
Long Term Stability		1000 Hours, +125°C (Note 3)		10		ppm/ $\sqrt{month}$

**Note 1:** Full-Scale Range (FSR) is 10V for unipolar mode, 20V ( $\pm 10V$ ) for bipolar mode.

**Note 2:** Using internal feedback and reference inverting resistors.

**Note 3:** Guaranteed by design, not 100% tested in production.



# ICL7135

## Precision 4½ Digit Single Chip A/D Converter

INTERFACE

Intersil

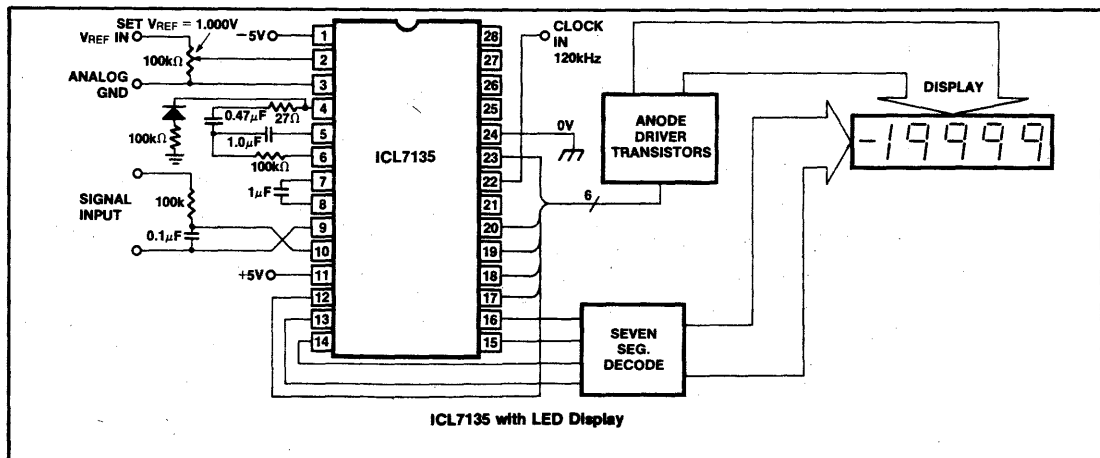
### FEATURES

- Accuracy guaranteed to  $\pm 1$  count over entire  $\pm 20,000$  counts (2.0000 volts full scale)
- Guaranteed zero reading for 0 volts input
- 1 pA typical input current
- True differential input
- True polarity at zero count for precise null detection
- Single reference voltage required
- Over-range and under-range signals available for auto-ranging capability
- All outputs TTL compatible
- Blinking display gives visual indication of over-range
- Six auxiliary inputs/outputs are available for interfacing to UARTs, microprocessors or other complex circuitry
- Multiplexed BCD output versatility

### GENERAL DESCRIPTION

The Intersil ICL7135 precision A/D converter, with its multiplexed BCD output and digit drivers, combines dual-slope conversion reliability with  $\pm 1$  in 20,000 count accuracy and is ideally suited for the visual display DVM/DPM market. The 2.0000V full scale capability, auto-zero and auto-polarity are combined with true ratiometric operation, almost ideal differential linearity and true differential input. All necessary active devices are contained on a single CMOS I.C., with the exception of display drivers, reference, and a clock.

The Intersil ICL7135 brings together an unprecedented combination of high accuracy, versatility, and true economy. High accuracy like auto-zero to less than  $10\mu\text{V}$ , zero drift of less than  $1\mu\text{V}/^\circ\text{C}$ , input bias current of 10 pA max., and rollover error of less than one count. The versatility of multiplexed BCD outputs is increased by the addition of several pins which allow it to operate in more sophisticated systems. These include  $\overline{\text{STROBE}}$ ,  $\overline{\text{OVERRANGE}}$ ,  $\overline{\text{UNDER-RANGE}}$ ,  $\overline{\text{RUN/HOLD}}$  and  $\overline{\text{BUSY}}$  lines, making it possible to interface the circuit to a microprocessor or UART.



ORDERING INFORMATION				PIN CONFIGURATION (Outline dwgs J1, P1)	
Part	Package	Temp. Range	Order Part #		
7135	28-Pin Cerdip	0°C to +70°C	ICL7135CJI		
7135	28-Pin Plastic DIP	0°C to +70°C	ICL7135CPI		
EV/ KIT	Evaluation Kit (PC Board, active, passive components)		ICL135EV/ KIT		

Pin	Function	Pin	Function
1	V <sup>-</sup>	28	UNDERRANGE
2	REFERENCE	27	OVERRANGE
3	ANALOG COMMON	26	STROBE
4	INT OUT	25	R/FI
5	AZ IN	24	DIGITAL GND
6	BUFF OUT	23	POL
7	REF. CAP. -	22	CLOCK IN
8	REF. CAP. +	21	BUSY
9	IN LO	20	(LSD) D1
10	IN HI	19	D2
11	V <sup>+</sup>	18	D3
12	(MSD) D5	17	D4
13	(LSB) B1	16	(MSB) B8
14	B2	15	B4

# ICL7135



## ABSOLUTE MAXIMUM RATINGS

Power Dissipation (Note 2)	Supply Voltage $V^+$ .....	+6V
Ceramic Package .....	$V^-$ .....	-9V
Plastic Package .....	Analog Input Voltage (either input) (Note 1) .....	$V^+$ to $V^-$
Operating Temperature .....	Reference Input Voltage (either input) .....	$V^+$ to $V^-$
Storage Temperature .....	Clock Input .....	Gnd to $V^+$
Lead Temperature (Soldering, 10 sec) .....		

**Note 1:** Input voltages may exceed the supply voltages provided the input current is limited to +100 $\mu$ A.  
**Note 2:** Dissipation rating assumes device is mounted with all leads soldered to printed circuit board.  
**\*COMMENT:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the devices. This is a stress rating only and functional operation of the devices at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ICL7135 ELECTRICAL CHARACTERISTICS (Note 1)

$V^+ = +5V$ ,  $V^- = -5V$ ,  $T_A = 25^\circ C$ , Clock Frequency Set for 3 Reading/Sec

	CHARACTERISTICS	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
ANALOG	Zero Input Reading		$V_{IN} = 0.0V$ Full Scale = 2.000V	-0.0000	$\pm 0.0000$	+0.0000	Digital Reading	
	Ratiometric Reading (2)		$V_{IN} \equiv V_{REF}$ Full Scale = 2.000V	+0.9998	+0.9999	+1.0000	Digital Reading	
	Linearity over $\pm$ Full Scale (error of reading from best straight line)		$-2V \leq V_{IN} \leq +2V$		0.5	1	Digital Count Error	
	Differential Linearity (difference between worse case step of adjacent counts and ideal step)		$-2V \leq V_{IN} \leq +2V$		.01		LSB	
	Rollover error (Difference in reading for equal positive & negative voltage near full scale)		$-V_{IN} \equiv +V_{IN} \approx 2V$		0.5	1	Digital Count Error	
	(Note 1) Noise (P-P value not exceeded 95% of time)	$e_n$	$V_{IN} = 0V$ Full scale = 2.000V		15		$\mu V$	
	(Note 2) Leakage Current at Input	$I_{ILK}$	$V_{IN} = 0V$		1	10	pA	
	Zero Reading Drift		$V_{IN} = 0V$ $0^\circ \leq T_A \leq 70^\circ C$		0.5	2	$\mu V/^\circ C$	
Scale Factor Temperature Coefficient (3)	TC	$V_{IN} = +2V$ $0 \leq T_A \leq 70^\circ C$ (ext. ref. 0 ppm/ $^\circ C$ )		2	5	ppm/ $^\circ C$		
DIGITAL	INPUTS	Clock In, Run/Hold, See Fig. 2	$V_{INH}$	2.8	2.2		V	
			$V_{INL}$		1.6	0.8	V	
			$I_{INL}$		0.02	0.1	mA	
			$I_{INH}$		0.1	10	$\mu A$	
	OUTPUTS	All Outputs B1, B2, B4, B8 D1, D2, D3, D4, D5 BUSY, STROBE, OVER-RANGE, UNDER-RANGE POLARITY	$V_{OL}$	$I_{OL} = 1.6ma$		0.25	0.40	V
			$V_{OH}$	$I_{OH} = -1mA$	2.4	4.2	V	
			$V_{OH}$	$I_{OH} = -10\mu A$	4.9	4.99	V	
	SUPPLY			$V^+$	+4	+5	+6	V
				$V^-$	-3	-5	-8	V
				$I^+$	$f_c = 0$		1.1	3.0
$I^-$				$f_c = 0$		0.8	3.0	mA
Power Dissipation Capacitance				CPD	vs. Clock Freq		40	
Clock	Clock Freq. (Note 4)		DC	2000	1200		kHz	

**Note 1:** Tested in 4-1/2 digit (20,000 count) circuit shown in Fig. 1, clock frequency 120kHz.  
**Note 2:** Tested with a low dielectric absorption integrating capacitor. See Component Selection Section.  
**Note 3:** The temperature range can be extended to +70 $^\circ C$  and beyond as long as the auto-zero and reference capacitors are increased to absorb the higher leakage of the ICL7135.  
**Note 4:** This specification relates to the clock frequency range over which the ICL7135 will correctly perform its various functions. See "Max Clock Frequency" below for limitations on the clock frequency range in a system.

**FEATURES**

- First-reading recovery from overrange gives immediate "OHMS" measurement
- Guaranteed zero reading for 0V input
- True polarity at zero for precise null detection
- 1pA typical input current
- True differential input and reference
- Direct LCD display drive — no external components required
- Pin compatible with the ICL7106, ICL7126
- Low noise — 15μVp-p without hysteresis or overrange hangover
- On-chip clock and reference
- Low power dissipation, guaranteed less than 1mW — gives 8,000 hours typical 9V battery life
- No additional active circuits required
- Evaluation Kit available (ICL7136EV/KIT)

**GENERAL DESCRIPTION**

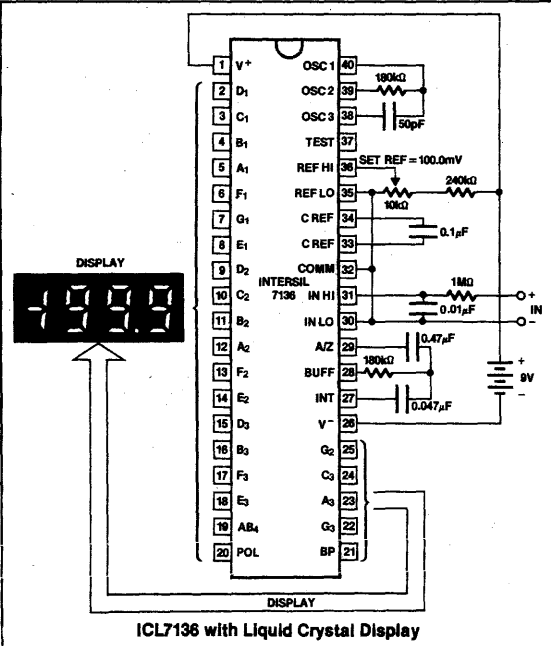
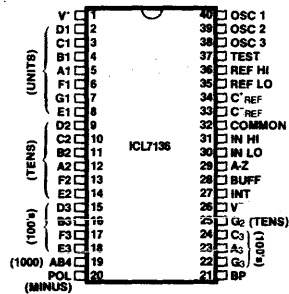
The Intersil ICL7136 is a high performance, very low power 3½-digit A/D converter. All the necessary active devices are contained on a single CMOS IC, including seven-segment decoders, display drivers, reference, and clock. The 7136 is designed to interface with a liquid crystal display (LCD) and includes a backplane drive. The supply current is under 100μA, ideally suited for 9V battery operation.

The 7136 brings together an unprecedented combination of high accuracy, versatility, and true economy. High accuracy, like auto-zero to less than 10μV, zero drift of less than 1μV/°C, input bias current of 10pA max., and rollover error of less than one count. The versatility of true differential input and reference is useful in all systems, but gives the designer an uncommon advantage when measuring load cells, strain gauges and other bridge-type transducers. And finally the true economy of single power supply operation allows a high performance panel meter to be built with the addition of only 7 passive components and a display.

The ICL7136 is an improved version of the ICL7126, eliminating the overrange hangover and hysteresis effects, and should be used in its place in all applications. It can also be used as a plug-in replacement for the ICL7106 in a wide variety of applications, changing only the passive components.

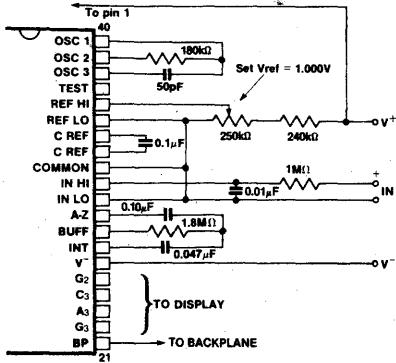
**ORDERING INFORMATION**

PART	PACKAGE	TEMPERATURE RANGE	ORDER PART NUMBER
7136	40-pin CERDIP	0°C to +70°C	ICL7136CJL
7136	40-pin Ceramic DIP	0°C to +70°C	ICL7136CDL
7136	40-pin Plastic DIP	0°C to +70°C	ICL7136CPL
7136 Kit	Evaluation Kits		ICL7136EV/KIT

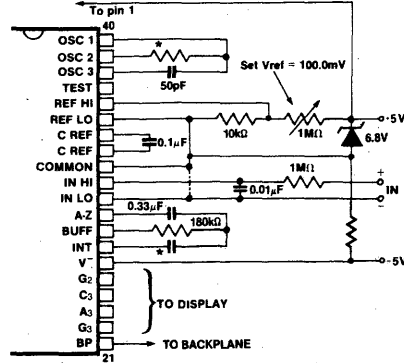
**PIN CONFIGURATION (Outline dwgs. DL, JL PL)**


**ICL7136**

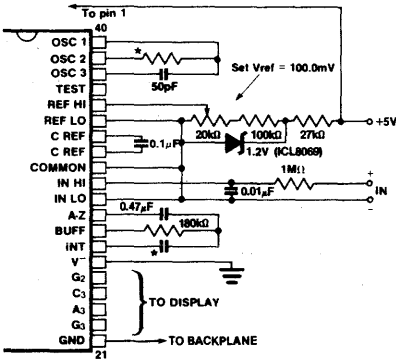
**TYPICAL APPLICATIONS (Continued)**



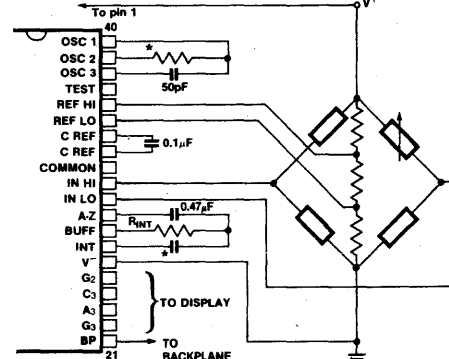
**Figure 11. Recommended Component Values for 2.000V Full-Scale, 3 Readings/Sec.** For 1 reading/sec, change  $C_{INT}$ ,  $R_{OSC}$  to values of Figure 10.



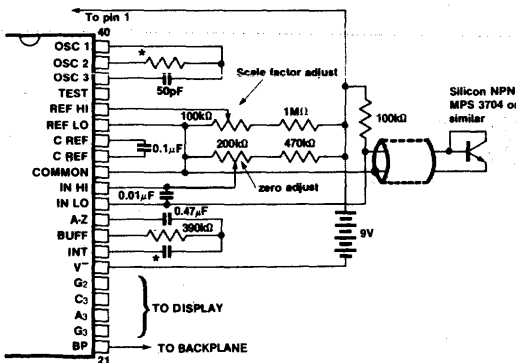
**Figure 12. 7136 with Zener Diode Reference.** Since low TC zeners have breakdown voltages ~6.8V, diode must be placed across the total supply (10V). As in the case of Figure 11, IN LO may be tied to COMMON.



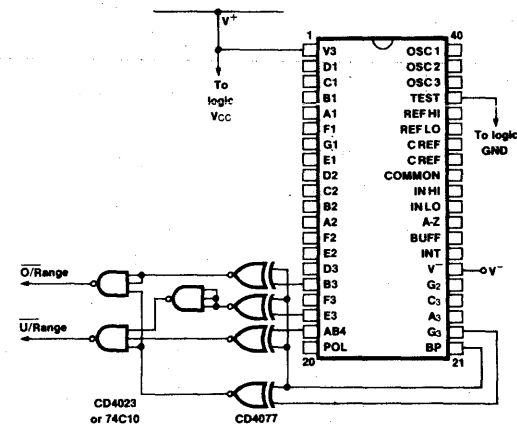
**Figure 13. 7136 Operated from Single +5V Supply.** An external reference must be used in this application, since the voltage between  $V^+$  and  $V^-$  is insufficient for correct operation of the internal reference.



**Figure 14. 7136 Measuring Ratioetric Values of Quad Load Cell.** The resistor values within the bridge are determined by the desired sensitivity.



**Figure 15. 7136 used as a Digital Centigrade Thermometer.** A silicon diode-connected transistor has a temperature coefficient of about  $-2mV/^\circ C$ . Calibration is achieved by placing the sensing transistor in ice water and adjusting the zeroing potentiometer for a 000.0 reading. The sensor should then be placed in boiling water and the scale-factor potentiometer adjusted for a 100.0 reading. See ICL8073/4 and AD590 data sheets for alternative circuits.



**Figure 16. Circuit for Developing Underrange and Overrange Signals from 7136 Outputs.**

\*Values depend on clock frequency. See Figures 9, 10, 11.



**PRELIMINARY**  
Specifications Subject To Change Without Notice

# ICL7145

## 16-Bit $\mu$ P-Compatible Multiplying D/A Converter

### FEATURES

- 16-bit resolution
- High linearity—0.003% FSR
- Microprocessor compatible with buffered inputs
- Bipolar application requires no external resistors
- Output current settling time 3 $\mu$ s max (1.0 $\mu$ s typ)
- Low linearity and gain temperature coefficients (1ppm/ $^{\circ}$ C typ)
- Low power dissipation
- Full four-quadrant multiplication
- Full temperature range operation

### GENERAL DESCRIPTION

The ICL7145 combines a four-quadrant multiplying DAC using thin film resistors and CMOS circuitry with an on-chip PROM-controlled correction circuit to achieve 0.003% linearity without laser trimming.

Microprocessor bus interfacing is eased by standard memory  $\overline{WR}$ ite cycle timing and control signal use. The input buffer register is loaded with the 16-bit input, and directly controls the output switches. The register is transparent if  $\overline{WR}$  and  $\overline{CS}$  are held low.

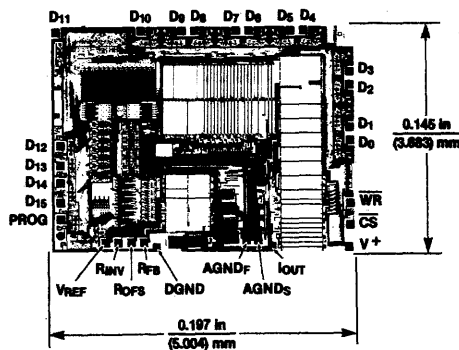
The ICL7145 is designed and programmed for bipolar operation. There is an offset resistor to the output with a reference input which should be connected to  $-V_{REF}$ , giving the DAC a true 2's complement input transfer function. Two extra resistors to facilitate the reference inversion are included on the chip, so that only an external op amp is needed.

### ORDERING INFORMATION

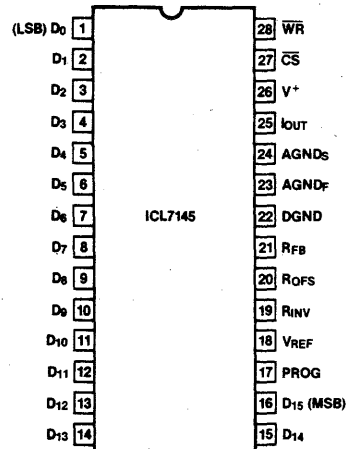
NON-LINEARITY	TEMPERATURE RANGE	
	0 $^{\circ}$ C TO +70 $^{\circ}$ C	-25 $^{\circ}$ C TO +85 $^{\circ}$ C
0.006%	ICL7145JCJI	ICL7145JJI
0.003%	ICL7145KCJI	ICL7145KIJI

Package: 28-pin CERDIP only

### CHIP TOPOGRAPHY



### PIN CONFIGURATION (outline dwg JI)





# ICL7145



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage  $V^+$  to DGND ..... -0.3V to 7.5V  
 $V_{REF}$ ,  $R_{OFS}$ ,  $R_{INV}$ ,  $R_{FB}$  to DGND .....  $\pm 25V$   
 Current in  $AGND_F$ ,  $AGND_S$  ..... 25mA  
 $D_N$ ,  $\overline{WR}$ ,  $\overline{CS}$ , PROG,  $I_{OUT}$ ,  
 $AGND_F$ ,  $AGND_S$  ..... -0.3V to  $V^+ + 0.3V$   
 Operating Temperature  
 ICL7145C ..... 0°C to +70°C  
 ICL7145I ..... -25°C to +85°C

Storage Temperature ..... -65°C to +150°C  
 Power Dissipation (Note 2) ..... 500mW  
 derate above 70°C @ 10mW/°C  
 Lead Temperature (soldering, 10 seconds) ..... 300°C

**Note 1:** All voltages with respect to DGND.

**Note 2:** Assumes all leads soldered or welded to printed circuit board.

*Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.*

## DC ELECTRICAL CHARACTERISTICS $V^+ = +5V$ , $V_{REF} = +5V$ , $T_A = +25^\circ C$ unless otherwise specified.

PARAMETER	SYMBOL	CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
Resolution			16			Bits
Non-Linearity	J	(Notes 3 and 4)			0.006	% FSR
	K				0.003	
Differential Non-Linearity		(Notes 3 and 4)		0.003		% FSR
Non-Linearity Temperature Coefficient		Operating Temperature Range		1		ppm/°C
Gain Error	J	(Notes 3 and 4)			0.04	% FSR
	K				0.02	
Gain Error Temperature Coefficient		Operating Temperature Range		1		ppm/°C
Zero Output Offset	$V_{OZ}$	$T_A = +25^\circ C$			10	mV
		Operating Temperature Range		10		
Power Supply Rejection Ratio	PSRR	$T_A = +25^\circ C$ , $V^+ = 5V \pm 10\%$		1	20	ppm/V
Output Current Settling Time				1	3	$\mu s$
Reference Input Resistance	$Z_{REF}$	$V_{REF}$	3		6	k $\Omega$
Output Capacitance	$C_{OUT}$	$D_N = \text{All } 0s$		110		pF
		$D_N = \text{All } 1s$		260		
Output Noise		Equivalent Johnson Resistance		7		k $\Omega$
Low State Input	$V_{INI}$	Operating Temperature Range			0.8	V
High State Input	$V_{INH}$	Operating Temperature Range	2.4			
Logic Input Current	$I_{LIN}$	$0 \leq V_{IN} \leq V^+$	-1.0		1.0	$\mu A$
Logic Input Capacitance	$C_{LIN}$			15		pF
Supply Voltage Range	$V^+$	Functional Operation	4.5		5.5	V
Supply Current	$I^+$	Excluding Ladder		0.5	1.2	mA

**Note 3:** Full-Scale Range (FSR) is 10V ( $\pm 5V$ ).

**Note 4:** Using internal feedback and reference inverting resistors.

## AC ELECTRICAL CHARACTERISTICS $V^+ = +5V$ , $T_A = +25^\circ C$ , see Timing Diagram.

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNIT
Chip Select-Write Set-Up Time	$t_{CWS}$				0	ns
Chip Select-Write Hold Time	$t_{CWH}$				0	
Write Pulse Width Low	$t_{WR}$				200	
Data-Write Set-Up Time	$t_{DWS}$				200	
Data-Write Hold Time	$t_{DWH}$				0	

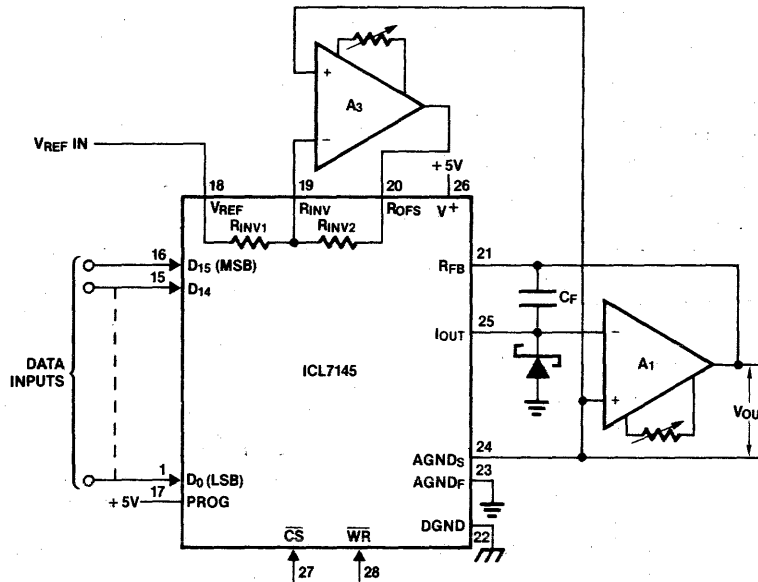


Figure 2. Bipolar Operation, Four-Quadrant

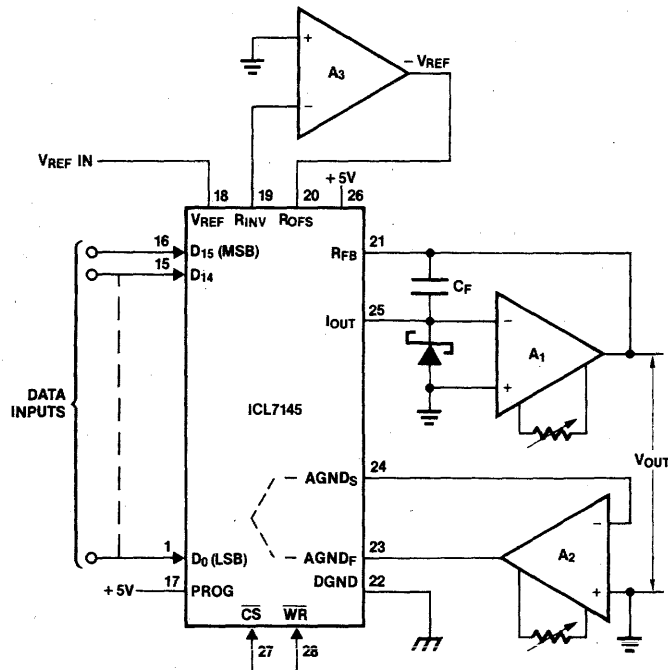


Figure 3. Operation with Forced Ground



# ICL7146 Complete 12-Bit Processor Compatible CMOS DAC

## FEATURES

- Low Impedance Voltage Output
- Double-Buffered Processor Interface
- Easy-To-Use Bipolar Offset
- Multiplying Capability
- On-Chip Trimmed Reference
- 7  $\mu$ s Settling Time
- No External Gain or Offset Adjustment Required
- Low Power Dissipation 50mW
- No Critical External Components

## ORDERING INFORMATION

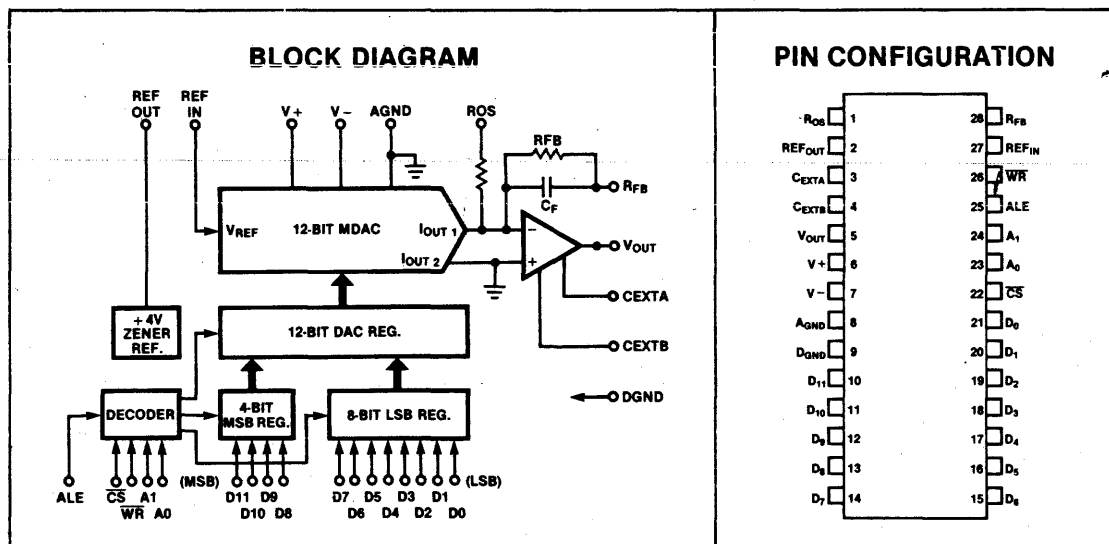
Part Number	Linearity	Temperature Range	Package
ICL7146LCJI	0.01%	0 to +70°C	CERDIP
ICL7146LIJI		-40°C to +85°C	
ICL7146KCJI	0.02%	0 to +70°C	
ICL7146KIJI		-40°C to +85°C	
ICL7146JCJI	0.05%	0 to +70°C	
ICL7146JIJI		-40°C to +85°C	

## GENERAL DESCRIPTION

The ICL7146 is the first of a series of complete 12-Bit CMOS DAC's. These DAC's feature all of the needed support circuitry to interface to processors and give a voltage output. Contained on the chip are two levels of latches for double buffers, a trimmed reference, a latch controller, and an output buffer amplifier. All devices are accurately trimmed for both gain and offset so that no external trimming is required.

CMOS circuitry is used to keep the power dissipation low, and with all devices contained on a single chip, significant board size reductions are possible. As an alternative to this, many more analog channels could be added to a board and still decrease power consumption. Intersil's patented autostabilized op amp construction eliminates drifts in the zero offset and provides a fast (7  $\mu$ s) settling time.

Processor interface is double-buffered with all 12-bits being brought out. The first level of latches is divided into 4 and 8 bit bytes with a 12 bit wide second buffer. Data can be directly entered into any of the three buffers or the buffers can be operated separately.



its output value. Neither of these conditions are acceptable in a wide variety of applications. Hence the need for double buffering.

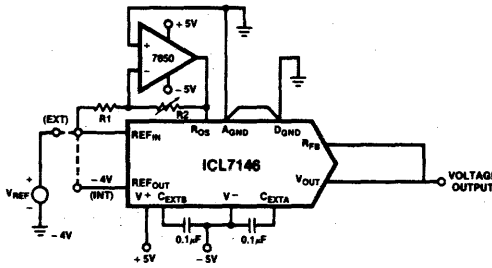
Buffer control is handled by a decoder to ease processor interface requirements. Operation of the decoder is shown in the truth table.

### TYPICAL APPLICATIONS Bipolar Output

Offset Binary Code Table

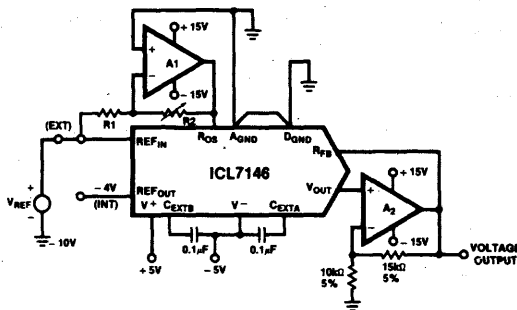
Binary Number In DAC Register		Analog Output, V <sub>OUT</sub>
MSB	LSB	
1111	1111 1111	$+V_{REF} \left( \frac{2047}{2048} \right)$
1000	0000 0001	$+V_{REF} \left( \frac{1}{2048} \right)$
1000	0000 0000	0V
0111	1111 1111	$-V_{REF} \left( \frac{1}{2048} \right)$
0000	0000 0000	$-V_{REF} \left( \frac{2048}{2048} \right)$

#### ± 4V BIPOLAR OUTPUT:



**NOTE 1:** A1 should be selected or trimmed for low offset voltage; R1 & R2 are 10kΩ resistors trimmed to a matching of 0.1% or better.

#### ± 10V BIPOLAR OUTPUT:



**NOTE 1:** A1 should be selected or trimmed for low offset voltage; R1 & R2 are 10kΩ resistors trimmed to a matching of 0.1% or better.

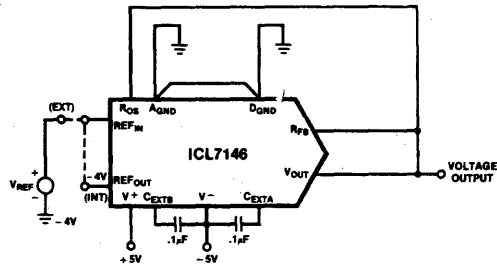
**NOTE 2:** A2 needs not to have a low offset voltage but it must be fast (>8MHz) to insure stability.

### TYPICAL APPLICATIONS Unipolar Output

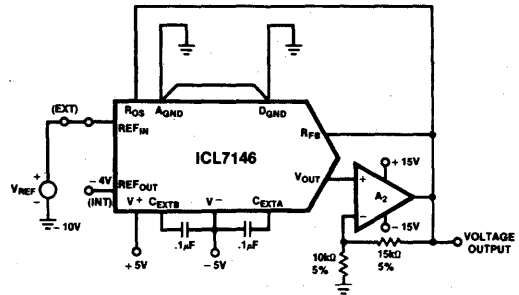
Code Table

Binary Number In DAC Register		Analog Output, V <sub>OUT</sub>
MSB	LSB	
1111	1111 1111	$-V_{REF} \left( \frac{4095}{4096} \right)$
1000	0000 0000	$-V_{REF} \left( \frac{2048}{4096} \right) = -1/2 V_{REF}$
0000	0000 0001	$-V_{REF} \left( \frac{1}{4096} \right)$
0000	0000 0000	0V

#### + 4V UNIPOLAR OUTPUT:



#### + 10V UNIPOLAR OUTPUT:



**NOTE 1:** A2 needs not to have a low offset voltage but it must be fast (>8MHz) to insure stability.

# ICL7146



## Absolute Maximum Ratings (Note 1)

REF <sub>IN</sub> , R <sub>FB</sub> , R <sub>OS</sub> .....	± 25V
V <sup>+</sup> .....	6.2V
V <sup>-</sup> .....	-9.0V
REF <sub>OUT</sub> , V <sub>OUT</sub> , C <sub>EXT</sub> ,	
AGND .....	V <sup>-</sup> - 0.3V to V <sup>+</sup> + 0.3V
Digital Inputs .....	V <sup>+</sup> + 0.3V to D <sub>GND</sub> - 0.3V
Storage Temperature Range .....	-65°C to +150°C

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

NOTE 1: All Voltages with Respect to D<sub>GND</sub>

## Operating Characteristics

V<sup>+</sup> = 5V, V<sup>-</sup> = -5V, V<sub>ref</sub> = 4.00V, T<sub>A</sub> = 25°C, R<sub>L</sub> = 20k, C<sub>L</sub> = 50pF

Parameter	Symbol	Test Conditions	Limits			Unit
			Min.	Typ.	Max.	
Resolution			12			Bits
Non Linearity	J K L				.05 .02 .01	% FSR % FSR % FSR
Differential Linearity	J K L	Guaranteed Monotonic		± 3/4 ± 1/2 ± 1/2	± 2 ± 1 ± 1	LSB
Gain Error				0.1	0.2	% FSR
Unipolar Zero Error				60	120	μV
Bipolar Zero Code Error		R <sub>FB</sub> Connected to V <sub>OUT</sub> R <sub>OS</sub> Connected to -V <sub>REF</sub>		0.025	0.05	% FSR
Positive Power Supply Rejection Ratio		V <sup>+</sup> = 4.5 to 5.5V External Reference		± 0.001	0.005	% FSR/ % V <sup>+</sup>
Negative Power Supply Rejection Ratio		V <sup>-</sup> = -4.5 to -5.5V External Reference		0	± 0.001	% FSR/ % V <sup>+</sup>
Voltage Setting Time (Note 1)		To 1/2 LSB		7	10	μs
Feedthrough Error		V <sub>REF</sub> = 8V P-P, 10 kHz Sine Wave			1	mV P-P
Reference Input Resistance		-55°C to 125°C	5	10	20	KΩ
Internal Reference Voltage			-4.04	-4.00	-3.96	V
Internal Reference Tempco				25	50	PPM of FSR per °C
Positive Supply Voltage Range	V <sup>+</sup>	Functional Operation, Internal or External Reference	4.5	5.0	5.5	V
Negative Supply Voltage Range	V <sup>-</sup>	Functional Operation, External Reference	-4.5	-5.0	-7.5	V
		Functional Operation, Internal Reference	-4.75	-5.0	-7.5	V
Output Voltage Range		R <sub>FB</sub> connected to V <sub>OUT</sub>		± 4		V
Output Drive Current			± 2			mA



**PRELIMINARY**  
Specifications Subject To Change Without Notice

# ICL7667

## Dual Power MOS Driver

### FEATURES

- 1.5A Peak Output Current
- Fast Rise and Fall Times  
— 40ns with 1000pF load
- Wide Supply Voltage Range  
—  $V_{CC} = 4.5$  to 20V
- Low Power Consumption  
— 4mW with inputs low  
— 120mW with inputs high
- TTL/CMOS Input Compatible Power Driver  
—  $R_{OUT} = 6\Omega$
- Direct Interface with Common Switching Regulators
- Pin Equivalent to DS0026/DS0056

### TYPICAL APPLICATIONS

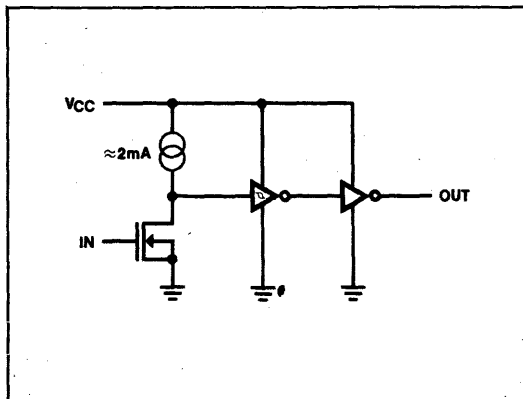
- Switching Power Supplies
- DC/DC Converters
- Motor Controllers

### ORDERING INFORMATION

Temperature Range	Package	Order Number
-55°C to +125°C	TO-99 Can	ICL7667MTV
	8-Pin Cerdip	ICL7667MJA
0 to +70°C	8-Pin Plastic	ICL7667CPA
	8-Pin Cerdip	ICL7667CJA
	TO-99 Can	ICL7667CJA
0 to +70°C	Dice	ICL7667C/D

(pin configuration for TV and PA packages also on this page)

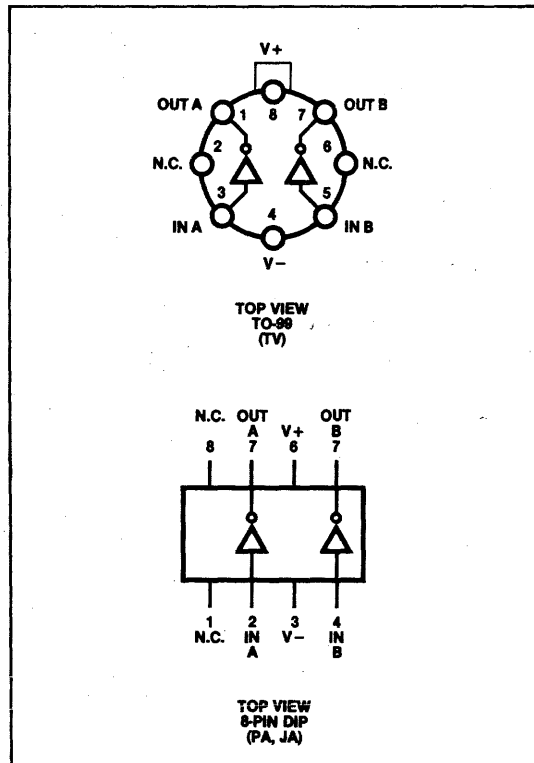
### BLOCK DIAGRAM



### GENERAL DESCRIPTION

The ICL7667 is a dual monolithic high-speed driver designed to convert TTL level signals into high current outputs at voltages up to 20V. Its high speed and 1.5A peak current output enable it to drive large capacitive loads with high slew rates and low propagation delays. With an output voltage swing only millivolts less than the supply voltage and a maximum supply voltage of 20V, the ICL7667 is well suited for driving power MOSFETs in high frequency switching regulators. The ICL7667's high current (1.5A peak) outputs minimize power losses in the power MOSFETs by rapidly charging and discharging the gate capacitances, while the ICL7667's inputs are TTL compatible and can be directly driven by common switching regulator IC's.

### PIN CONFIGURATION



**ICL7667****ABSOLUTE MAXIMUM RATINGS**

Supply Voltage .....	22V
Input Voltage .....	22V to $(V^- - 0.3V)$
Peak Output Current .....	1.5A
Package Dissipation, $T_A = 25^\circ C$ .....	500mW
Linear Derating Factors	
TO-99	Plastic
6.7mW/ $^\circ C$	5.6mW/ $^\circ C$
above $50^\circ C$	above $36^\circ C$
Cerdip	6.7mW/ $^\circ C$
above $50^\circ C$	
Storage Temperature .....	$-65^\circ C$ to $+150^\circ C$

Lead Temperature (Soldering, 10 seconds) .....	$300^\circ C$
Operating Temperature Range	
C Series .....	0 to $+70^\circ C$
M Series .....	$-55^\circ C$ to $+125^\circ C$

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**DC OPERATING CHARACTERISTICS**

Test Conditions:  $V_{CC} = 4.5$  to  $20V$ ,  $T_A = +25^\circ C$  unless otherwise noted.

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNITS
			MIN	TYP	MAX	
Logic 1 Input Voltage	$V_{IH}$		2.4	2.0		V
Logic 0 Input Voltage	$V_{IL}$			1.5	0.8	V
Input Current	$I_{IL}$	$0 < V_{IN} < V_{CC}$	-1	0	1	$\mu A$
Output Voltage High	$V_{OH}$	No Load	$V_{CC} - 0.05$	$V_{CC}$		V
Output Voltage Low	$V_{OL}$	No Load		0	0.05	V
Output Resistance	$R_{OUT}$	$V_{IN} = V_{IL}$ $I_{OUT} = -10mA$ $V_{CC} = 20V$		6	20	$\Omega$
Output Resistance	$R_{OUT}$	$V_{IN} = V_{IH}$ $I_{OUT} = 10mA$ $V_{CC} = 20V$		6	20	$\Omega$
Power Supply Current	$I_{CC}$	$V_{IN} = 3V$ (both inputs)		4	6	mA
Power Supply Current	$I_{CC}$	$V_{IN} = 0V$ (both inputs)		150	400	$\mu A$

**AC OPERATING CHARACTERISTICS**

Test Conditions:  $V_{CC} = 20V$ ,  $T_A = +25^\circ C$ , unless otherwise noted.

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNITS
			MIN	TYP	MAX	
Delay Time	$T_{D2}$	Figure 1		50	75	ns
Delay Time	$T_{D2}$	Figure 2		50	75	ns
Rise Time	$T_R$	Figure 1		25	35	ns
Rise Time	$T_R$	Figure 2		35	50	ns
Fall Time	$T_F$	Figure 1		30	40	ns
Fall Time	$T_F$	Figure 2		40	55	ns
Delay Time	$T_{D1}$	Figure 1		20	35	ns
Delay Time	$T_{D1}$	Figure 2		20	35	ns

### FEATURES

- High frequency operation — 10MHz guaranteed
- Easy to use oscillator — requires only a quartz crystal and two capacitors
- Bipolar, MOS and CMOS compatibility
- High output drive capability — 5 x TTL fanout with 10ns rise and fall times
- Low power — 50mW at 10MHz
- Choice of two output frequencies — osc., and osc. ÷8 frequencies
- Disable control for both outputs
- Wide industrial temperature range — -20°C to +85°C
- All inputs fully protected — circuits may be handled without any special precautions

### GENERAL DESCRIPTION

The Intersil ICM7209 is a versatile CMOS clock generator capable of driving a number of 5 volt systems with a variety of input requirements. When used to drive up to 5 TTL gates, the typical rise and fall times are 10ns.

The ICM7209 consists of an oscillator, a buffered output equal to the oscillator frequency and a second buffered output having an output frequency one-eighth that of the oscillator. The guaranteed maximum oscillator frequency is 10MHz. Connecting the DISABLE terminal to the negative supply forces the ÷8 output into the '0' state and the output 1 into the '1' state.

<h4>SCHEMATIC DIAGRAM</h4> <p>*ZENER VOLTAGE IS TYPICALLY 6.3 VOLTS</p>	<h4>CHIP TOPOGRAPHY</h4>
<h4>ORDERING INFORMATION</h4> <p>Order Devices by Following Part Number ICM7209 I PA Order Dice by Following Part Number ICM7209/D</p>	<h4>PIN CONFIGURATION (OUTLINE DRAWING PA)</h4> <p>TOP VIEW</p> <p>Pin 1 is designated by either a dot or a notch.</p>



# ICM7209



## ABSOLUTE MAXIMUM RATINGS

Power Dissipation (25°C)	300mW
Supply Voltage	6 V
Output Voltages	Equal to or less than supply
Input Voltages	Equal to or less than supply
Storage Temp.	-55°C to +125°C
Operating Temp. Range	-20°C to +85°C
Lead Temp. (Soldering, 10 seconds)	300°C

**NOTE:** Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## TYPICAL OPERATING CHARACTERISTICS

**TEST CONDITIONS:** V+ = 5V ±10%, test circuit, f<sub>osc</sub> = 10MHz, T<sub>A</sub> = 25°C unless otherwise specified.

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP.	MAX	UNITS
Supply Current	I*	Note 1 No Load		11	20	mA
Disable Input Capacitance	C <sub>D</sub>				5	pF
Disable Input Leakage	I <sub>ILK</sub>	Either '1' or '0' state			±10	μA
Output Low State	V <sub>OL</sub>	Either OUT 1 or OUT ÷8 simulated 5 x TTL loads			0.4	V
Output High State	V <sub>OH</sub>	Either OUT 1 or OUT ÷8 simulated 5 x TTL loads	4.0	4.9		
Output Rise Time (Note 3)	t <sub>r</sub>	Either OUT 1 or OUT ÷8 simulated 5 x TTL loads		10	25	ns
Output Fall Time (Note 3)	t <sub>f</sub>	Either OUT 1 or OUT ÷8 simulated 5 x TTL loads		10	25	
Minimum OSC Frequency for ÷8 Output	f <sub>osc</sub>	Note 2	2			MHz
Output ÷8 duty cycle		Any operating frequency Low state : High state		7:9		
Oscillator Transconductance	gm		80	200		μmho

**NOTE 1:** The power dissipation is a function of the oscillator frequency (1st ORDER EFFECT see curve) but is also effected to a small extent by the oscillator tank components.

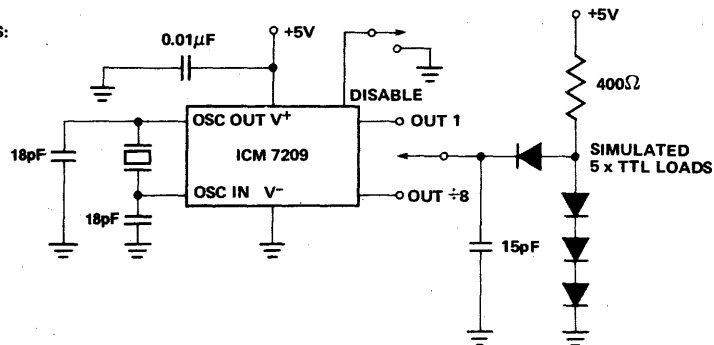
**NOTE 2:** The ÷8 circuitry uses a dynamic scheme. As with any dynamic system, information or data is stored on very small nodal capacitances instead of latches (static systems) and there is a lower cutoff frequency of operation. Dynamic dividers are used in the ICM7209 to significantly improve high frequency performance and to decrease power consumption.

**NOTE 3:** Rise and fall times are defined between the output levels of 0.5 and 2.4 volts.

## TEST CIRCUIT

### CRYSTAL PARAMETERS:

C<sub>M</sub> = 5mpF  
R<sub>S</sub> = 15 ohms  
C<sub>O</sub> = 3pF  
C<sub>L</sub> = 10pF  
f = 10 MHz





# ICM7243

## 8-Character 14-/16-Segment Alphanumeric LED Display Driver

### FEATURES

- 14- and 16-segment fonts with decimal point
- Mask programmable for other font-sets up to 64 characters
- Microprocessor compatible
- Directly drives small common cathode displays
- Cascadable without additional hardware
- Standby feature turns display off; puts chip in low power mode
- Serial entry or random entry of data into display
- Single +5V operation
- Character and segment drivers, all MUX scan circuitry, 8 x 6 static memory and 64-character ASCII font generator included on-chip

### GENERAL DESCRIPTION

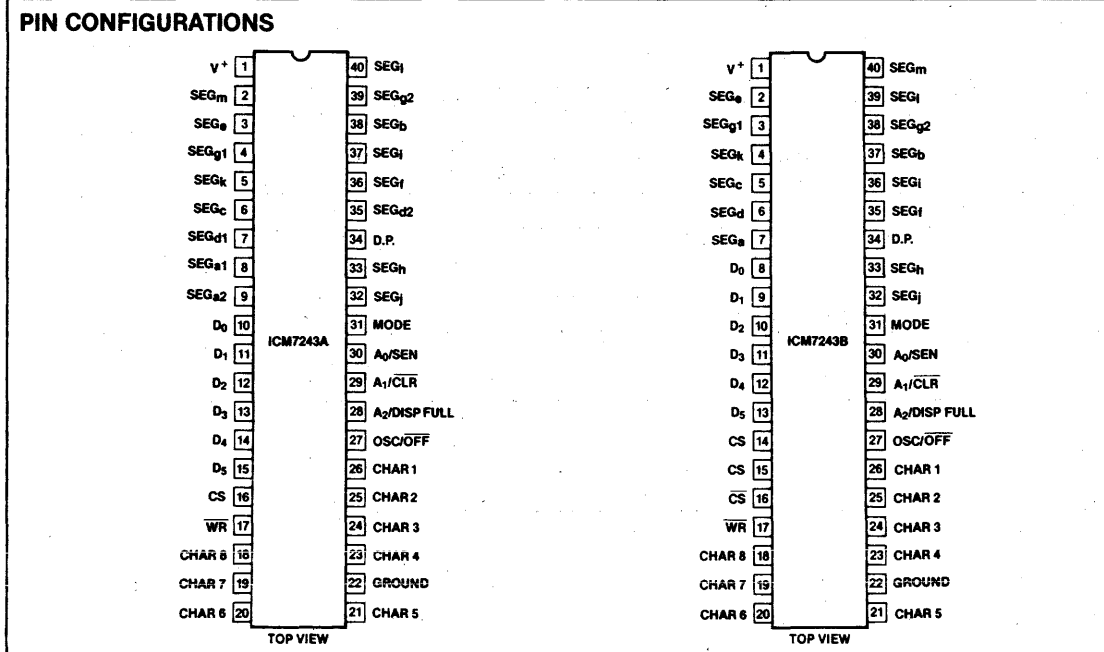
The ICM7243 is an 8-character alphanumeric display driver and controller which provides all the circuitry required to interface a microprocessor or digital system to a 14- or 16-segment display. It is primarily intended for use in microprocessor systems, where it offloads the processor and minimizes hardware and software overhead. Incorporated on-chip are a 64-character ASCII decoder, an 8x6 memory, the high power character and segment drivers, and the multiplex scan circuitry.

Six-bit ASCII data to be displayed is written into the memory directly from the microprocessor data bus. Data location depends upon the selection of either **Serial** (MODE = 1) or **Random** (MODE = 0). In the **Serial Access** mode the first entry is stored in the lowest location and displayed in the "left-most" character position. Each subsequent entry is automatically stored in the next higher location and displayed to the immediate "right" of the previous entry. A **DISPlay FULL** signal is provided after 8 entries; this signal can be used for cascading. A **CLear** pin is provided to clear the memory and reset the location counter. The **Random Access** mode allows the processor to select the memory address and display digit for each input word.

The character multiplex scan runs whenever data is not being entered. It scans the memory and **CHARacter** drivers, and ensures that the decoding from memory to display is done in the proper sequence. Intercharacter blanking is provided to avoid display ghosting.

### ORDERING INFORMATION

Part Number	Display Segments	Package	Order Number
ICM7243A	16 + d.p.	40 Pin CERDIP	ICM7243AIJL
ICM7243B	14 + d.p.	40 Pin CERDIP	ICM7243BIJL
ICM7243B EVKIT	Kit with Display		ICM7243B EVKIT



# ICM7243



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage	.....6V	Power Dissipation	.....1W
CHARacter Output Current	.....300mA	Operating Temperature Range	.....-20°C to +85°C
SEGment Output Current	.....30mA	Storage Temperature Range	.....-55°C to +125°C
Input Voltage (Any Terminal)	.....(V <sup>+</sup> +0.3V) to -0.3V		

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## OPERATING CHARACTERISTICS

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
Supply Voltage	V <sup>+</sup>		4.75	5.0	5.25	V
Operating Supply Current	I <sup>+</sup> <sub>OP</sub>	V <sup>+</sup> = 5.25V, 10 Segments ON, All 8 Characters		180		mA
Quiescent Supply Current	I <sub>Q</sub> <sup>+</sup>	V <sup>+</sup> = 5.25V, OSC/OFF Pin < 1V		30	250	μA
Input High Voltage	V <sub>IH</sub>		2			V
Input Low Voltage	V <sub>IL</sub>				0.8	V
Input Current	I <sub>IN</sub>	V <sup>+</sup> = 5.25V, V <sub>IH</sub> = 5V V <sub>IL</sub> = 0V	-1		+1	μA
CHARacter Drive Current	I <sub>CHAR</sub>	V <sup>+</sup> = 5V, V <sub>OUT</sub> = 1V	140	190		mA
CHARacter Leakage Current	I <sub>CHLK</sub>					μA
SEGment Drive Current	I <sub>SEG</sub>	V <sup>+</sup> = 5V, V <sub>OUT</sub> = 2.5V	14	19		mA
SEGment Leakage Current	I <sub>SLK</sub>			0.01		μA
DISPlay FULL Output Low	V <sub>OL</sub>	I <sub>OL</sub> = 1.6mA			0.4	V
DISPlay FULL Output High	V <sub>OH</sub>	I <sub>IH</sub> = 100μA	2.4			V
Display Scan Rate	f <sub>ds</sub>			400		Hz

## AC CHARACTERISTICS (Drive levels 0.4V and 2.4V, timing measured at 0.8V and 2.0V)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
WR, CLeaR Pulse Width Low	t <sub>WPl</sub>		250			ns
WR, CLeaR Pulse Width High	t <sub>WPh</sub>		250			
Data Hold Time	t <sub>Dh</sub>		0	-20		
Data Setup Time	t <sub>Ds</sub>		250	150		
Address, SEN, MODE Hold Time	t <sub>Ah</sub>		125	80		
Address, SEN, MODE Setup Time	t <sub>As</sub>		-20			
CS, CS Setup Time	t <sub>Cs</sub>		0			
Pulse Transition Time	t <sub>t</sub>				100	

## CAPACITANCE

SYMBOL	TEST	MIN	TYP	MAX	UNIT
C <sub>IN</sub>	Input Capacitance				pF
C <sub>O</sub>	Output Capacitance				pF

# ICM7243

APPLICATIONS (Continued)

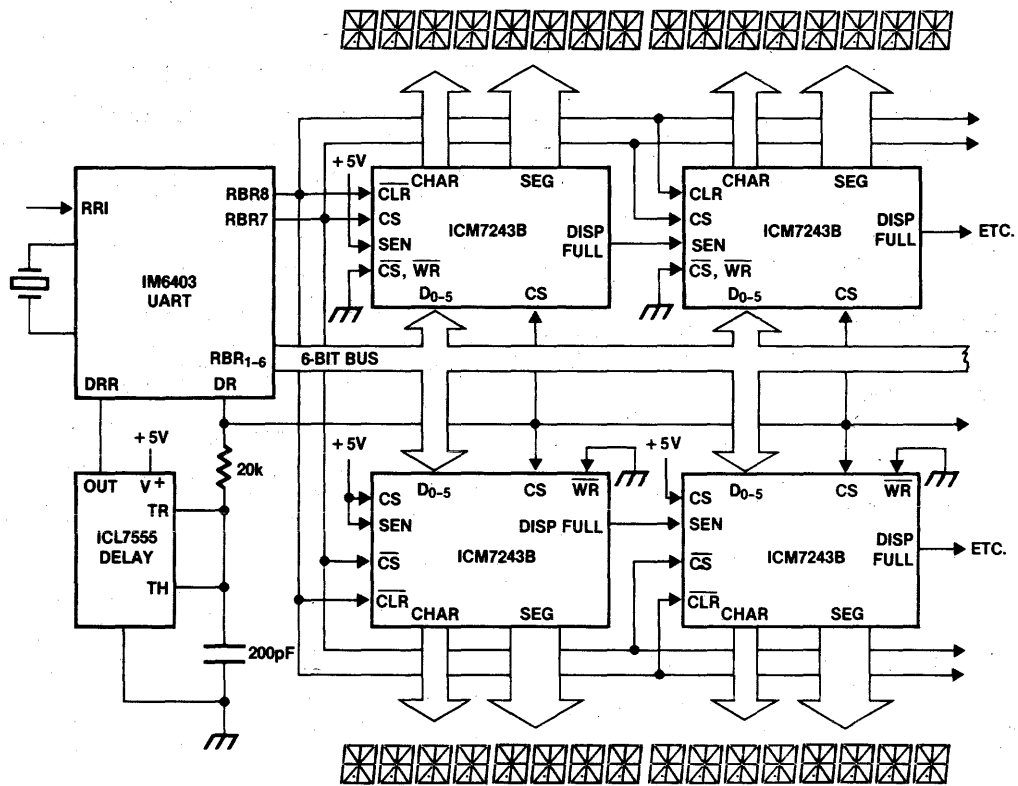


Figure 2. Driving Two Rows of Characters from a Serial Input. UART converts data stream to parallel bytes. Bit 7 of each word sets which row data will be entered into. Bit 8 will blank and reset whole display if low. Each MODE pin should be tied high. ICM7243A can also be used, with inverter on RBR7 for one row.

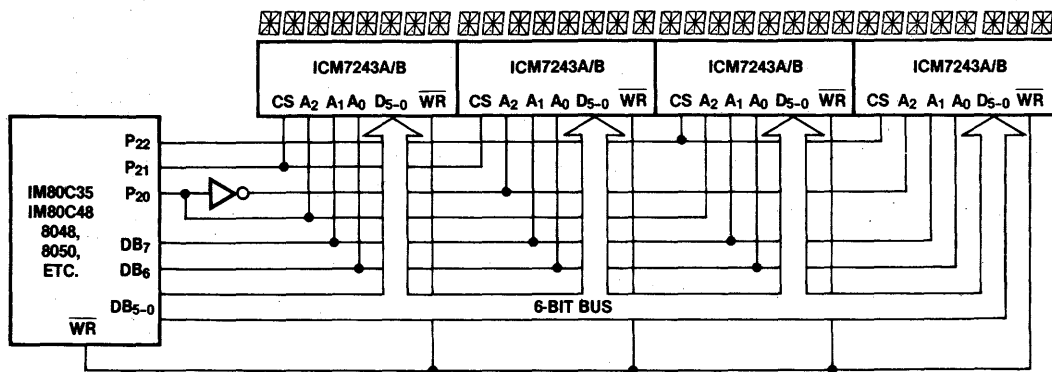


Figure 3. Random Access 32-Character Display in MCS-48 system. One port line controls A<sub>2</sub>, other two are CS lines. 8-bit data bus drives 6 data and 2 address lines. MODE should be Grounded on each part.



# ICM7281 LCD Column Driver

## FEATURES

- LCD Dot Matrix Column Driver
- 40 High Voltage LCD Column Drive Outputs For Up to 8 5xN Characters per IC
- Easy Interface
  - Serial Input Shift Register With parallel latch and carry outputs
- Directly Compatible with ICM7280 Row Driver
  - Up to 10 ICM7281's can be driven by an ICM7280 with no external components
- Low Resistance Outputs
  - Can drive both columns and rows of LCD graphics displays
- Will Drive 1.5V Threshold LCDs with Only Single 5V Supply
  - Can drive up to 4.5V threshold LCDs with 15V V<sub>DISP</sub>

## GENERAL DESCRIPTION

The ICM7281 LCD Dot Matrix Column Driver is designed to convert a serial data stream into drive signals for a multiplexed dot matrix LCD. Easily cascadable, up to 10 ICM7281's can be driven by one ICM7280 Intelligent Row Driver to make an 80 character dot matrix display. The ICM7281 also serves as both a Row Driver and Column Driver in LCD dot matrix graphics displays. The low output resistance and the 15V drive capability make it well suited for graphics displays with up to 256 x 256 dots (with 10pF/dot capacitance).

The ICM7281 consists of a 40 bit shift register, a 40 bit latch and 40 level-shifters/drivers. The 4 display drive voltages are generated externally, usually by a Row Driver. A serial data interface is used to minimize the number of pins needed for digital interfacing. Two data Carry Outputs are included for cascading several ICM7281's to drive large LCD displays.

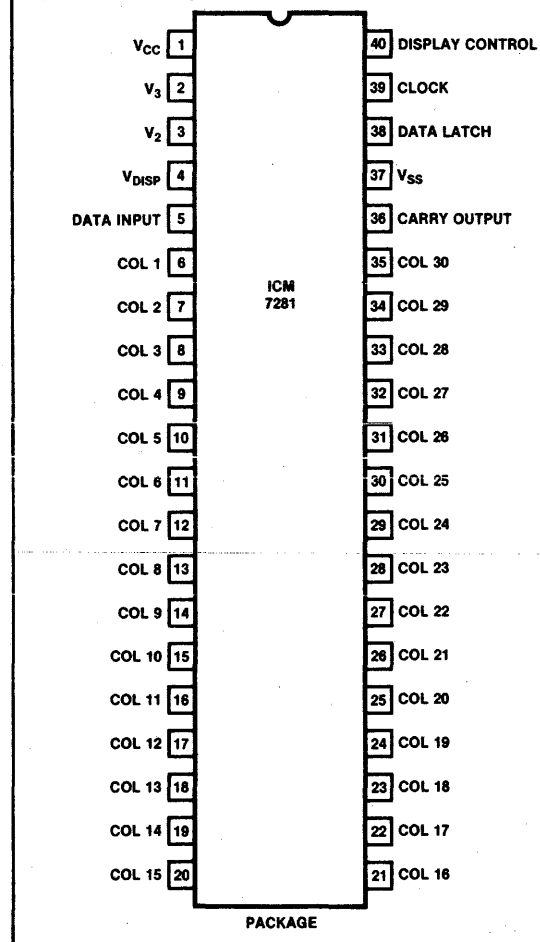
## ORDERING INFORMATION

No. Of Columns	Package	Order Number
30	40 Pin Plastic	ICM7281IPL
40	Dice	ICM7281/D
40	52-64 Pin Plastic Flatpack	—

## TYPICAL APPLICATIONS

- Column Drivers for Dot Matrix Alpha-numeric Displays using ICM7280 Row Driver
- Row and Column Drivers for LCD Dot Matrix Graphics Displays
- Segment Driver for LCD Bargraphs and Annunciators
- Serial Input I/O Expander

## PIN CONFIGURATIONS (Outline dwg. PL)



### FEATURES

- Switches Greater Than 20Vpp Signals With  $\pm 15V$  Supplies
- Quiescent Current Less Than  $1\mu A$
- Overvoltage Protection to  $\pm 25V$
- Break-Before-Make Switching  $t_{off}$  200 nsec,  $t_{on}$  300 nsec Typical
- $T^2L$ , DTL, CMOS, PMOS Compatible
- Non-Latching With Supply Turn-Off
- Low  $r_{DS(on)}$  -  $35\Omega$
- New DPDT & 4PST Configurations
- Complete Monolithic Construction IH5040 through IH5047

### FUNCTIONAL DIAGRAM

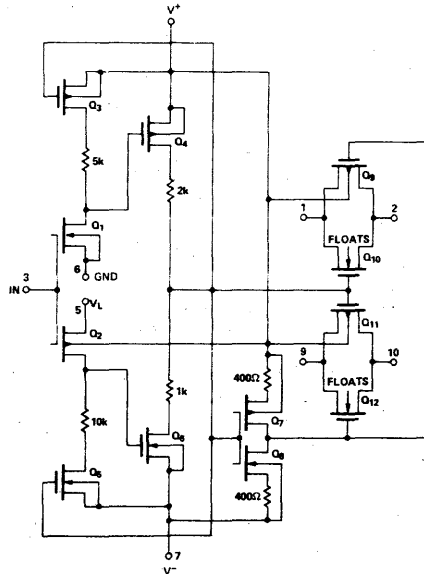
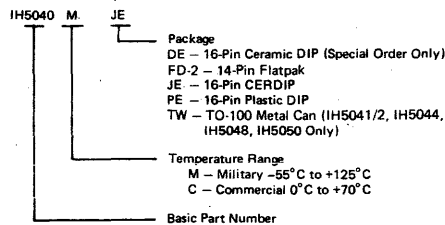


FIGURE 1. TYPICAL DRIVER, GATE - IH5042

### ORDERING INFORMATION



### GENERAL DESCRIPTION

The IH5040 family of solid state analog gates are designed using an improved, high voltage CMOS monolithic technology. These devices provide ease-of-use and performance advantages not previously available from solid state switches. This improved CMOS technology provides input overvoltage capability to  $\pm 25$  volts without damage to the device, and destructive latch-up of solid state analog gates has been eliminated. Early CMOS gates were destroyed when power supplies were removed with an input signal present. The IH5040 CMOS technology has eliminated this serious systems problem.

Key performance advantages of the 5040 series are TTL compatibility and ultra low-power operation. The quiescent current requirement is less than  $1\mu A$ . Also designed into the 5040 is guaranteed Break-Before-Make switching, which is accomplished by extending the  $t_{on}$  time (300 nsec TYP.) so that it exceeds  $t_{off}$  time (200 nsec TYP.). This insures that an ON channel will be turned OFF before an OFF channel can turn ON. This eliminates the need for external logic required to avoid channel to channel shorting during switching.

Many of the 5040 series improve upon and are pin-for-pin and electrical replacements for other solid state switches.

### FUNCTIONAL DESCRIPTION

INTERSIL PART NO.	TYPE	$r_{DS(on)}$	PIN/FUNCTIONAL EQUIVALENT (Note 1)
IH5040	SPST	75Ω	
IH5041	Dual SPST	75Ω	
IH5042	SPDT	75Ω	DG 188AA/BA
IH5043	Dual SPDT	75Ω	DG 191AP/BP
IH5044	DPST	75Ω	
IH5045	Dual DPST	75Ω	DG 185AP/BP
IH5046	DPDT	75Ω	
IH5047	4PST	75Ω	
IH5048 Dual	SPST	35Ω	
IH5049 Dual	DPST	35Ω	DG 184AP/BP
IH5050	SPDT	35Ω	DG 187AA/BA
IH5051 Dual	SPDT	35Ω	DG 190AP/BP

NOTE 1. See Switching State diagrams for applicable package equivalency.

Pin and functional equivalent monolithic versions of the DG181, DG182, DG187 and DG188 are available. See data sheet for this and also IH181 to IH191.

# IH5040-IH5051 Family



## ABSOLUTE MAXIMUM RATINGS

Current (Any Terminal)	< 30mA
Storage Temperature	-65°C to +150°C
Operating Temperature	-55°C to +125°C
Power Dissipation	450mW
(All Leads Soldered to a P.C. Board)	
Derate 6mW/°C Above 70°C	
Lead Temperature (Soldering, 10 sec)	300°C

V <sup>+</sup> -V <sup>-</sup>	< 33V
V <sup>+</sup> -V <sub>D</sub>	< 30V
V <sub>D</sub> -V <sup>-</sup>	< 30V
V <sub>D</sub> -V <sub>S</sub>	< ±22V
V <sub>L</sub> -V <sup>-</sup>	< 33V
V <sub>L</sub> -V <sub>IN</sub>	< 30V
V <sub>L</sub> -GND	< 20V
V <sub>IN</sub> -GND	< 20V

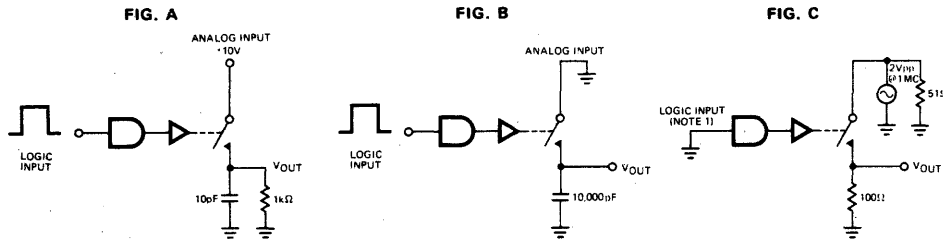
Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS (@ 25°C, V<sup>+</sup> = +15 V, V<sup>-</sup> = -15 V, V<sub>L</sub> = +5 V)

PER CHANNEL		MIN./MAX. LIMITS						UNITS	TEST CONDITIONS
		MILITARY			COMMERCIAL				
SYMBOL	CHARACTERISTIC	-55°C	+25°C	+125°C	0	+25°C	+70°C		
I <sub>IN(ON)</sub>	Input Logic Current	1	1	1	1	1	1	μA	V <sub>IN</sub> = 2.4 V Note 1
I <sub>IN(OFF)</sub>	Input Logic Current	1	1	1	1	1	1	μA	V <sub>IN</sub> = 0.8 V Note 1
r <sub>DS(on)</sub>	Drain-Source On Resistance	75(35)	75(35)	150(60)	80(45)	80(45)	130(45)	Ω	(IH5048 Thru IH5051) I <sub>S</sub> = 10 mA V <sub>ANALOG</sub> = -10 V to +10 V
Δr <sub>DS(ON)</sub>	Channel to Channel r <sub>DS(ON)</sub> Match		25 (15) (typ)			30(15) (typ)		Ω	(IH5048 thru IH5051)
V <sub>ANALOG</sub>	Min. Analog Signal Handling Capability		±11(±10)			±10(±10)		V	
I <sub>D(OFF)</sub>	Switch OFF Leakage Current	1(1)	1(1)	100(100)	5(5)	5(5)	100(100)	nA	V <sub>ANALOG</sub> = -10 V to +10 V (IH5048 thru IH5051)
I <sub>D(ON)</sub>	Switch ON Leakage Current	2(2)	2(2)	200(200)	10(10)	10(10)	100(200)	nA	V <sub>D</sub> = V <sub>S</sub> = -10 V to +10 V (IH5048 thru IH5051)
t <sub>on</sub>	Switch "ON" Time		500(250)			500(300)		ns	R <sub>L</sub> = 1 kΩ, V <sub>ANALOG</sub> = -10 V to +10 V See Fig. A
t <sub>off</sub>	Switch "OFF" Time		250(150)			250(150)		ns	R <sub>L</sub> = 1 kΩ, V <sub>ANALOG</sub> = -10 V to +10 V See Fig. A (IH5048 thru IH5051)
Q <sub>(INJ.)</sub>	Charge Injection		15 (10)			20 (10)		mV	See Fig. B (IH5048 thru IH5051)
OIRR	Min. Off Isolation Rejection Ratio		54			50		dB	f = 1 MHz, R <sub>L</sub> = 100Ω, C <sub>L</sub> ≤ 5 pF See Fig. C, (Note 1)
I <sub>Q<sup>+</sup></sub>	+ Power Supply Quiescent Current	1	1	10	10	10	100	μA	
I <sub>Q<sup>-</sup></sub>	- Power Supply Quiescent Current	1	1	10	10	10	100	μA	V <sup>+</sup> = +15 V, V <sup>-</sup> = -15 V, V <sub>L</sub> = +5 V
I <sub>LQ</sub>	+5 V Supply Quiescent Current	1	1	10	10	10	100	μA	V <sub>L</sub> = +5 V
I <sub>GND</sub>	Gnd Supply Quiescent Current	1	1	10	10	10	100	μA	
CCRR	Min. Channel to Channel Cross Coupling Rejection Ratio		54			50		dB	One Channel Off; Any Other Channel Switches as per Fig. E (Note 1)

Note 1: Not tested in production.

## TEST CIRCUITS



**NOTE 1:** Some channels are turned on by high "1" logic inputs and other channels are turned on by low "0" inputs; however 0.8V to 2.4V describes the min. range for switching properly. Refer to logic diagrams to see absolute value of logic input required to produce "ON" or "OFF" state.

# IH5040-IH5051 Family



## SWITCHING STATE DIAGRAMS

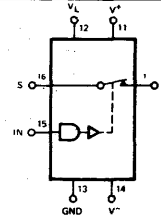
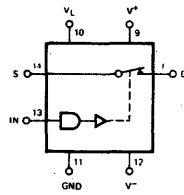
SWITCH STATES  
ARE FOR LOGIC "1" INPUT

(OUTLINE DWG  
FD-2)

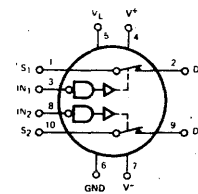
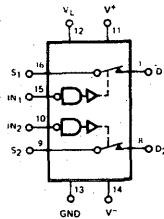
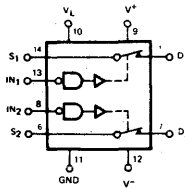
(OUTLINE DWGS  
DE, JE, PE)

(OUTLINE DWG TO-100)

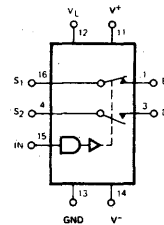
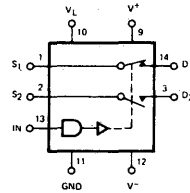
**SPST**  
IH5040 ( $r_{DS(on)} < 75\Omega$ )



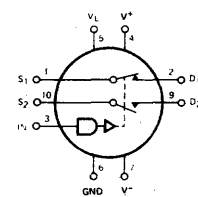
**DUAL SPST**  
IH5041 ( $r_{DS(on)} < 75\Omega$ )



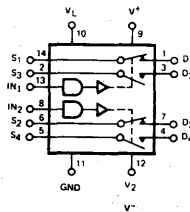
**SPDT**  
IH5042 ( $r_{DS(on)} < 75\Omega$ )



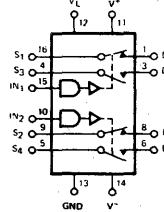
(DG188 EQUIVALENT)



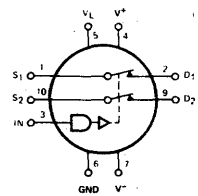
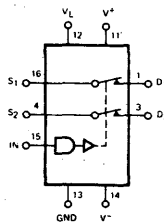
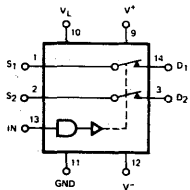
**DUAL SPDT**  
IH5043 ( $r_{DS(on)} < 75\Omega$ )



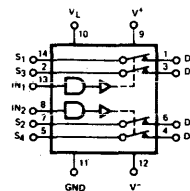
(DG191 EQUIVALENT)



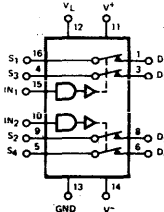
**DPST**  
IH5044 ( $r_{DS(on)} < 75\Omega$ )



**DUAL DPST**  
IH5045 ( $r_{DS(on)} < 75\Omega$ )



(DG185 EQUIVALENT)





# IH5040-IH5051 Family



## SWITCHING STATE DIAGRAMS (Cont.)

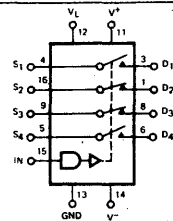
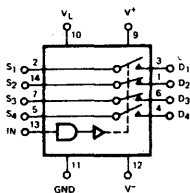
SWITCH STATES ARE FOR LOGIC "1" INPUT

FLAT PACKAGE (FD-2)

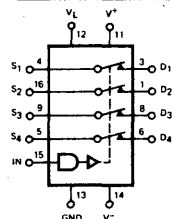
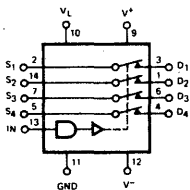
DIP (DE) PACKAGE

TO-100

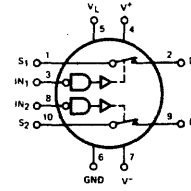
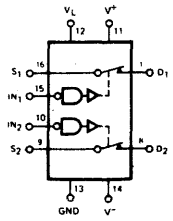
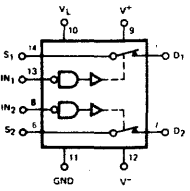
**DPDT**  
IH5046 ( $r_{DS(ON)} < 75\Omega$ )



**4PST**  
IH5047 ( $r_{DS(ON)} < 75\Omega$ )

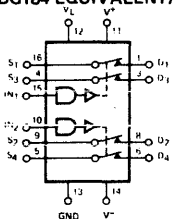
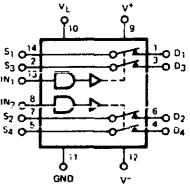


**DUAL SPST**  
IH5048 ( $r_{DS(ON)} < 35\Omega$ )



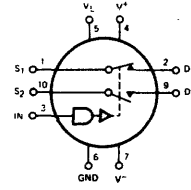
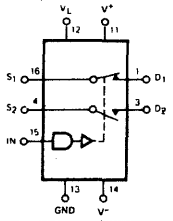
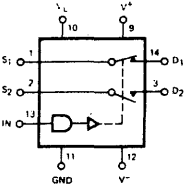
**DUAL DPST**  
IH5049 ( $r_{DS(ON)} < 35\Omega$ )

(DG184 EQUIVALENT)



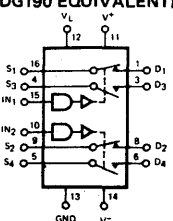
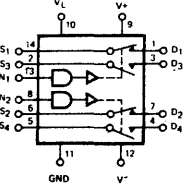
**SPDT**  
IH5050 ( $r_{DS(ON)} < 35\Omega$ )

(DG187 EQUIVALENT)



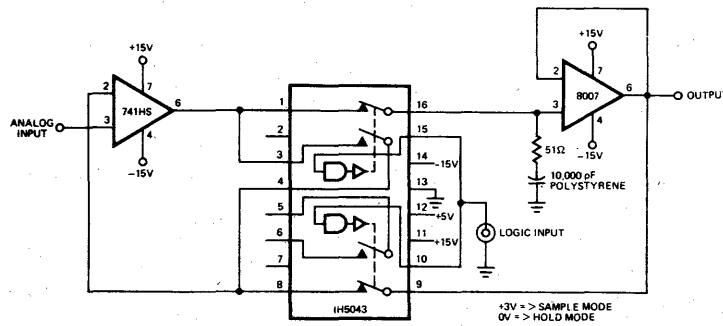
**DUAL SPDT**  
IH5051 ( $r_{DS(ON)} < 35\Omega$ )

(DG190 EQUIVALENT)



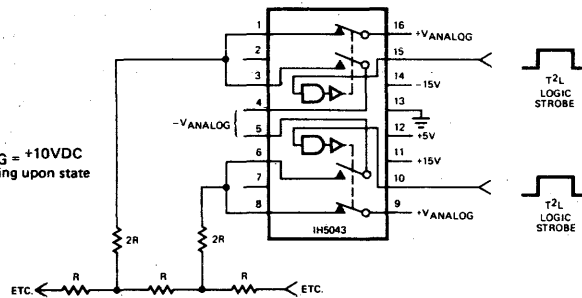
## APPLICATIONS

IMPROVED SAMPLE & HOLD  
USING IH5043

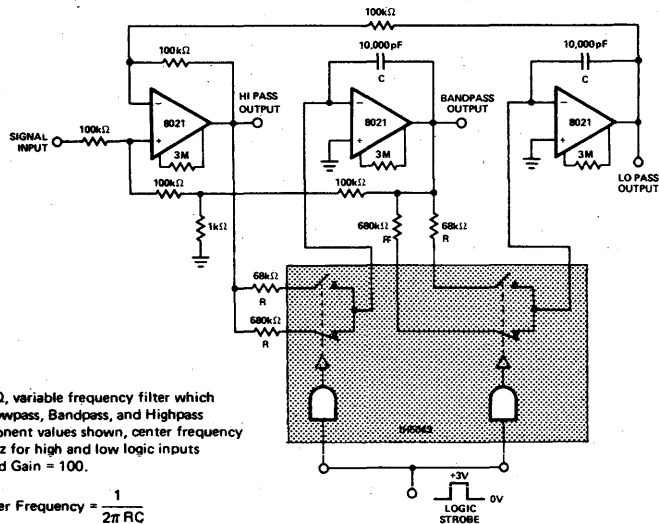


USING THE CMOS SWITCH TO DRIVE  
AN R/2R LADDER NETWORK (2 LEGS)

**EXAMPLE:** If  $-V_{ANALOG} = -10VDC$  and  $+V_{ANALOG} = +10VDC$  then Ladder Legs are switched between  $\pm 10VDC$ , depending upon state of Logic Strobe.



DIGITALLY TUNED  
LOW POWER ACTIVE FILTER



Constant gain, constant Q, variable frequency filter which provides simultaneous Lowpass, Bandpass, and Highpass outputs. With the component values shown, center frequency will be 235Hz and 23.5Hz for high and low logic inputs respectively, Q = 100, and Gain = 100.

$$f_n = \text{Center Frequency} = \frac{1}{2\pi RC}$$



# IH5108

## 8-Channel Fault Protected CMOS Analog Multiplexer

### FEATURES

- Ultra low leakage —  $I_{D(off)} \leq 100\text{pA}$
- Power supply quiescent current less than 1mA
- $\pm 13\text{V}$  analog signal range
- No SCR latchup
- Break-before-make switching
- Pin compatible with DG508, HI508 and AD7508
- All channels OFF ( $I_{ILK} \leq 100\text{nA}$ ) when power OFF, for analog signals up to  $\pm 25\text{V}$
- Any channel turns OFF ( $I_{ILK} \leq 100\text{nA}$ ) if input exceeds supply rails by up to  $\pm 25\text{V}$ . Throughput always  $< \pm 14\text{V}$  ( $\pm 15\text{V}$  supplies)
- TTL and CMOS compatible binary Address and ENable inputs

### GENERAL DESCRIPTION

The IH5108 is a dielectrically isolated CMOS monolithic analog multiplexer, designed as a plug-in replacement for the DG508 and similar devices, but adding fault protection to the standard performance. A unique serial MOSFET switch ensures that an OFF channel will remain OFF when the input exceeds the supply rails by up to  $\pm 25\text{V}$ , even with the supply voltage at zero. Further, an ON channel will be limited to a throughput of about 1.5V less than the supply rails, thus affording protection to any following circuitry such as op amps, D/A converters, etc. Cross talk onto "good" channels is also prevented.

A binary 3-bit address code together with the ENable input allows selection of any one channel or none at all. These 4 inputs are all TTL compatible for easy logic interface; the ENable input also facilitates MUX expansion and cascading.

#### FUNCTIONAL DIAGRAM

ADDRESS DECODE

#### DECODE TRUTH TABLE

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH
X	X	X	0	NONE
0	0	0	1	1
0	0	1	1	2
0	1	0	1	3
0	1	1	1	4
1	0	0	1	5
1	0	1	1	6
1	1	0	1	7
1	1	1	1	8

A<sub>0</sub>, A<sub>1</sub>, A<sub>2</sub>, EN  
 Logic "1" = V<sub>AH</sub> ≥ 2.4V  
 Logic "0" = V<sub>AL</sub> ≤ 0.8V

---

A<sub>0</sub> A<sub>1</sub> A<sub>2</sub> EN (ENABLE INPUT)

3 LINE BINARY ADDRESS INPUTS  
(1 0 1) AND EN HI  
ABOVE EXAMPLE SHOWS CHANNEL 6 TURNED ON

#### PIN CONFIGURATION (outline dwg JE, PE)

TOP VIEW

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#### ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH5108MJE	-55°C to +125°C	16 pin Cerdip
IH5108IJE	-20°C to +85°C	16 pin Cerdip
IH5108CPE	0°C to 70°C	16 pin plastic DIP

# IH5108



## ABSOLUTE MAXIMUM RATINGS

$V_{IN}(A, EN)$ to Ground	-15V to 15V
$V_S$ or $V_D$ to $V^+$	+25V, -40V
$V_S$ or $V_D$ to $V^-$	-25V, +40V
$V^+$ to Ground	16V
$V^-$ to Ground	-16V
Current (Any Terminal)	20mA

Operating Temperature	-55 to 125°C
Storage Temperature	-65 to 150°C
Power Dissipation (Package)*	1200mW

\* All leads soldered or welded to PC board. Derate 10mW/°C above 70°C.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS $V^+ = 15V, V^- = -15V, V_{EN} = 2.4V$ , unless otherwise specified.

CHARACTERISTIC	MEASURED TERMINAL	NO TESTS PER TEMP	TYP 25°C	MAX LIMITS						UNIT	TEST CONDITIONS		
				M SUFFIX			I/C SUFFIX						
				-55°C	25°C	125°C	-20°C/0°C	25°C	85°C/70°C				
SWITCH	$r_{DS(on)}$	S to D	8	700	900	900	1200	1200	1200	1800	$\Omega$	$V_D = 10V, I_S = -1.0mA$ $V_D = -10V, I_S = -1.0mA$ Sequence each switch on $V_{AL} = 0.8V, V_{AH} = 2.4V$	
			8	500	900	900	1200	1200	1200	1800			
	$\Delta r_{DS(on)}$			5					10		10	%	$\Delta r_{DS(on)} = \frac{r_{DS(on)max} - r_{DS(on)min}}{r_{DS(on)avg}} V_S = \pm 10V$
	$I_S(off)$	S	8	0.002		0.05	50		0.1	50	nA	$V_S = 10V, V_D = -10V$ $V_S = -10V, V_D = 10V$ $V_D = 10V, V_S = -10V$ $V_D = -10V, V_S = 10V$ $V_{EN} = 0$	
	8		0.002		0.05	50		0.1	50				
$I_D(off)$	D	1	0.03		0.1	100		0.2	100	nA	$V_{S(A1)} = V_D = 10V$ Sequence each switch on $V_{S(A1)} = V_D = -10V$ $V_{AL} = 0.8V, V_{AH} = 2.4V$		
1		0.03		0.1	100		0.2	100					
$I_D(on)$	D	8	0.1		0.2	100		0.4	100	nA	$V_{S(A1)} = V_D = 10V$ Sequence each switch on $V_{S(A1)} = V_D = -10V$ $V_{AL} = 0.8V, V_{AH} = 2.4V$		
8		0.1		0.2	100		0.4	100					
FAULT	$I_S$ with Power OFF	S	8	1		100	1000	50	50	5000	nA	$V^+ = V^- = 0V, V_S = \pm 25V, V_{EN} = V_O = 0V, A_0, A_1, A_2 = 0V$ or 5V $V_S = \pm 25V, V_D = \mp 10V$ Sequence each switch	
	$I_S(off)$ with Overvoltage (Note 1)	S	8	1		2000	5000		5000	5000			
IN	$I_{EN(on)} I_{A(on)}$ or $I_{EN(off)} I_{A(off)}$	$A_0, A_1, EN$	4	.01		-10	-30		-10	-30	$\mu A$	$V_A = 2.4V$ or 0V	
	4		.01		10	30		10	30	$V_A = 15V$ or 0V			
DYNAMIC	$t_{transition}$	D		0.3		1					$\mu s$	See Figure 1	
	$t_{open}$	D		0.2								See Figure 2	
	$t_{on(EN)}$	D		0.6		1.5						See Figure 3	
	$t_{off(EN)}$	D		0.4		1							
	$t_{on-t_{off}}$ Break-Before-Make Delay Settling Time	D	8	50		25			10		ns	$V_{EN} = +5V, A_0, A_1, A_2$ Strobed $V_{IN} = \pm 10V$ , Figure 4	
	"OFF" Isolation	D		60							dB	$V_{EN} = 0, R_L = 200\Omega, C_L = 3pF, V_S = 3 VRMS, f = 500 KHz$	
$C_{S(off)}$	S		5							pF	$V_S = 0$ $V_D = 0$ $V_S = 0, V_D = 0$		
$C_{D(off)}$	D		25								$V_{EN} = 0V, f = 140 KHz$ to 1 MHz		
$C_{DS(off)}$	D to S		1										
SUPPLY	Supply Current	+	$V^+$	1	500	900	750	600	1000		$\mu A$	All $V_A, V_{EN} = 0$ or 5V	
	Current	-	$V^-$	1	500	900	750	600	1000				

Note 1. Readings taken 400ms after the overvoltage occurs.



# IH5140 Family High Level CMOS Analog Gates

## FEATURES

- Super fast break before make switching  
 $t_{on}$  80ns typ,  $t_{off}$  50ns typ (SPST switches)
- Power supply currents less than  $1\mu A$
- OFF leakages less than 100pA @ 25°C guaranteed
- Non-latching with supply turn-off
- Single monolithic CMOS chip
- Plug-in replacements for IH5040 family and part of the DG180 family to upgrade speed and leakage
- Greater than 1MHz toggle rate
- Switches greater than 20Vp-p signals with  $\pm 15V$  supplies
- T<sup>2</sup>L, CMOS direct compatibility

## GENERAL DESCRIPTION

The IH5140 Family of CMOS monolithic switches utilizes Intersil's latch-free junction isolated processing to build the fastest switches now available. These switches can be toggled at a rate of greater than 1 MHz with super fast  $t_{on}$  times (80ns typical) and faster  $t_{off}$  times (50ns typical), guaranteeing break before make switching. This family of switches therefore combines the speed of the hybrid FET DG 180 Family with the reliability and low power consumption of a monolithic CMOS construction.

OFF leakages are guaranteed to be less than 100pA at 25°C. No quiescent power is dissipated in either the ON or the OFF state of the switch. Maximum power supply current is  $1\mu A$  from any supply and typical quiescent currents are in the 10nA range which makes these devices ideal for portable equipment and military applications.

The IH5140 Family is completely compatible with TTL (5V) logic, TTL open collector logic and CMOS logic gates. It is pin compatible with Intersil's IH5040 Family and part of the DG180/190 Family as shown in the switching state diagrams.

## ORDERING INFORMATION

Order Part Number	Function	Package	Temperature Range
IH5140 MJE	SPST	16 Pin CERDIP	-55°C to 125°C
IH5140 CJE	SPST	16 Pin CERDIP	0°C to 70°C
IH5140 CPE	SPST	16 Pin Plastic DIP	0°C to 70°C
IH5140 MFD	SPST	14 Pin Flat Pack	-55°C to 125°C
IH5141 MJE	Dual SPST	16 Pin CERDIP	-55°C to 125°C
IH5141 CJE	Dual SPST	16 Pin CERDIP	0°C to 70°C
IH5141 CPE	Dual SPST	16 Pin Plastic DIP	0°C to 70°C
IH5141 MFD	Dual SPST	14 Pin Flat Pack	-55°C to 125°C
IH5141 CTW	Dual SPST	TO-100	0°C to 70°C
IH5141 MTW	Dual SPST	TO-100	-55°C to 125°C
IH5142 MJE	SPDT	16 Pin CERDIP	-55°C to 125°C
IH5142 CJE	SPDT	16 Pin CERDIP	0°C to 70°C
IH5142 CPE	SPDT	16 Pin Plastic DIP	0°C to 70°C
IH5142 MFD	SPDT	14 Pin Flat Pack	-55°C to 125°C
IH5142 CTW	SPDT	TO-100	0°C to 70°C
IH5142 MTW	SPDT	TO-100	-55°C to 125°C
IH5143 MJE	Dual SPDT	16 Pin CERDIP	-55°C to 125°C
IH5143 CJE	Dual SPDT	16 Pin CERDIP	0°C to 70°C
IH5143 CPE	Dual SPDT	16 Pin Plastic DIP	0°C to 70°C
IH5143 MFD	Dual SPDT	14 Pin Flat Pack	-55°C to 125°C
IH5144 MJE	DPST	16 Pin CERDIP	-55°C to 125°C
IH5144 CJE	DPST	16 Pin CERDIP	0°C to 70°C
IH5144 CPE	DPST	16 Pin Plastic DIP	0°C to 70°C
IH5144 MFD	DPST	14 Pin Flat Pack	-55°C to 125°C
IH5144 CTW	DPST	TO-100	0°C to 70°C
IH5144 MTW	DPST	TO-100	-55°C to 125°C
IH5145 MJE	Dual DPST	16 Pin CERDIP	-55°C to 125°C
IH5145 CJE	Dual DPST	16 Pin CERDIP	0°C to 70°C
IH5145 CPE	Dual DPST	16 Pin Plastic DIP	0°C to 70°C
IH5145 MFD	Dual DPST	14 Pin Flat Pack	-55°C to 125°C

Note:  
1. Ceramic (side braze) devices also available; consult factory.  
2. MIL temp range parts also available with MIL-STD-883 processing.

## FUNCTIONAL DIAGRAM

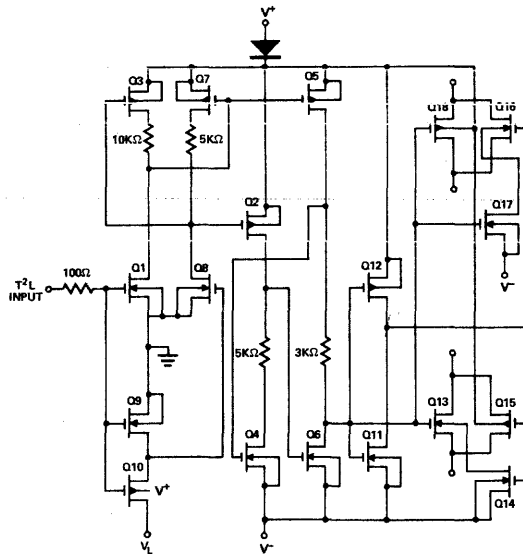
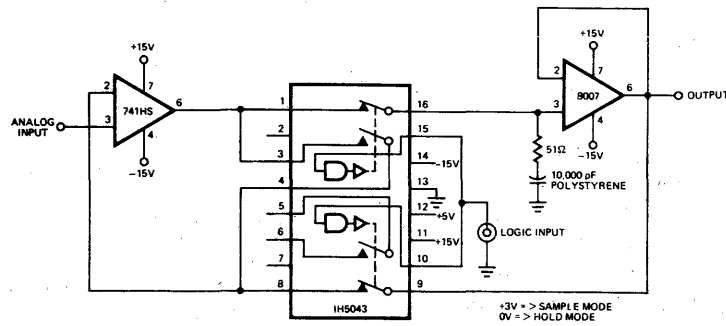


FIGURE 1. Typical Driver/Gate — IH5142

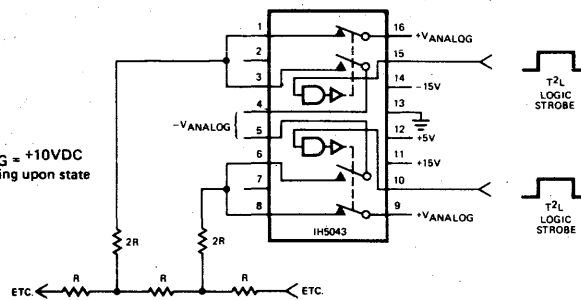
## APPLICATIONS

IMPROVED SAMPLE & HOLD  
USING IH5043

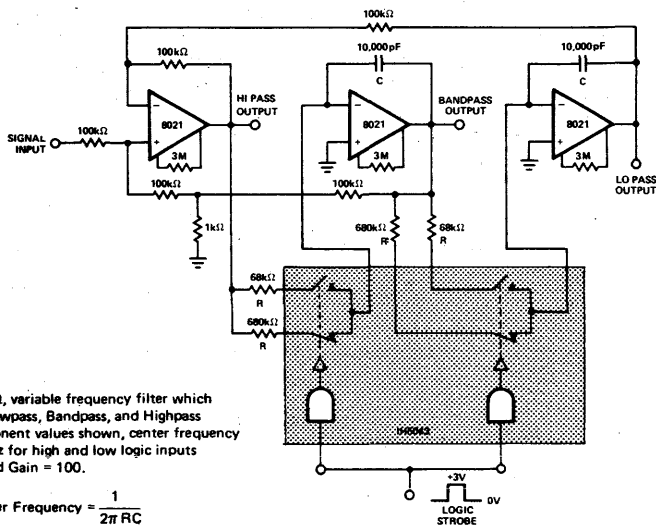


USING THE CMOS SWITCH TO DRIVE  
AN R/2R LADDER NETWORK (2 LEGS)

**EXAMPLE:** If  $-V_{ANALOG} = -10VDC$  and  $+V_{ANALOG} = +10VDC$  then Ladder Legs are switched between  $\pm 10VDC$ , depending upon state of Logic Strobe.



DIGITALLY TUNED  
LOW POWER ACTIVE FILTER



Constant gain, constant Q, variable frequency filter which provides simultaneous Lowpass, Bandpass, and Highpass outputs. With the component values shown, center frequency will be 235Hz and 23.5Hz for high and low logic inputs respectively,  $Q = 100$ , and Gain = 100.

$$f_n = \text{Center Frequency} = \frac{1}{2\pi RC}$$



# IH5108

## 8-Channel Fault Protected CMOS Analog Multiplexer

### FEATURES

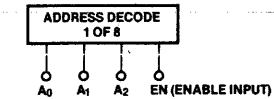
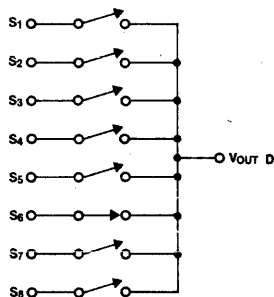
- Ultra low leakage —  $I_{D(off)} \leq 100\text{pA}$
- Power supply quiescent current less than 1mA
- $\pm 13\text{V}$  analog signal range
- No SCR latchup
- Break-before-make switching
- Pin compatible with DG508, HI508 and AD7508
- All channels OFF ( $I_{ILK} \leq 100\text{nA}$ ) when power OFF, for analog signals up to  $\pm 25\text{V}$
- Any channel turns OFF ( $I_{ILK} \leq 100\text{nA}$ ) if input exceeds supply rails by up to  $\pm 25\text{V}$ . Throughput always  $< \pm 14\text{V}$  ( $\pm 15\text{V}$  supplies)
- TTL and CMOS compatible binary Address and ENable inputs

### GENERAL DESCRIPTION

The IH5108 is a dielectrically isolated CMOS monolithic analog multiplexer, designed as a plug-in replacement for the DG508 and similar devices, but adding fault protection to the standard performance. A unique serial MOSFET switch ensures that an OFF channel will remain OFF when the input exceeds the supply rails by up to  $\pm 25\text{V}$ , even with the supply voltage at zero. Further, an ON channel will be limited to a throughput of about 1.5V less than the supply rails, thus affording protection to any following circuitry such as op amps, D/A converters, etc. Cross talk onto "good" channels is also prevented.

A binary 3-bit address code together with the ENable input allows selection of any one channel or none at all. These 4 inputs are all TTL compatible for easy logic interface; the ENable input also facilitates MUX expansion and cascading.

### FUNCTIONAL DIAGRAM



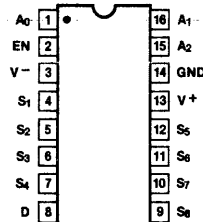
3 LINE BINARY ADDRESS INPUTS (1 0 1) AND EN HI ABOVE EXAMPLE SHOWS CHANNEL 6 TURNED ON

### DECODE TRUTH TABLE

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH
X	X	X	0	NONE
0	0	0	1	1
0	0	1	1	2
0	1	0	1	3
0	1	1	1	4
1	0	0	1	5
1	0	1	1	6
1	1	0	1	7
1	1	1	1	8

A<sub>0</sub>, A<sub>1</sub>, A<sub>2</sub>, EN  
 Logic "1" = V<sub>AH</sub> ≥ 2.4V  
 Logic "0" = V<sub>AL</sub> ≤ 0.8V

### PIN CONFIGURATION (outline dwg JE, PE)



TOP VIEW

### ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH5108MJE	-55°C to +125°C	16 pin CERDIP
IH5108IJE	-20°C to +85°C	16 pin CERDIP
IH5108CPE	0°C to 70°C	16 pin plastic DIP

# IH5140-IH5145 Family



## ABSOLUTE MAXIMUM RATINGS

Current (Any Terminal) ..... < 30 mA  
 Storage Temperature ..... -65°C to +150°C  
 Operating Temperature ..... -55°C to +125°C  
 Power Dissipation ..... 450 mW  
 (All Leads Soldered to a P.C. Board)  
 Derate 6 mW/°C Above 70°C  
 Lead Temperature (Soldering 10 sec.) .. 300°C

$V^+ - V^-$  < 33V  
 $V^+ - V_D$  < 30V  
 $V_D - V^-$  < 30V  
 $V_D - V_S$  < ±22V  
 $V_L - V^-$  < 33V  
 $V_L - V_{IN}$  < 30V  
 $V_L$  < 20V  
 $V_{IN}$  < 20V

NOTE: Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS (@ 25°C, $V^+ = +15V$ , $V^- = -15V$ , $V_L = +5V$ )

PER CHANNEL		MIN./MAX. LIMITS						UNITS	TEST CONDITIONS
		MILITARY			COMMERCIAL				
SYMBOL	CHARACTERISTIC	-55°C	+25°C	+125°C	0	+25°C	+70°C		
$I_{INH}$	Input Logic Current	1	1	1	1	1	1	μA	$V_{IN} = 2.4 V$ Note 1
$I_{INL}$	Input Logic Current	1	1	1	1	1	1	μA	$V_{IN} = 0.8 V$ Note 1
$r_{DS(on)}$	Drain—Source On Resistance	50	50	75	75	75	100	Ω	$I_S = -10 mA$ $V_{ANALOG} = -10 V$ to +10 V
$\Delta r_{DS(on)}$	Channel to Channel $r_{DS(on)}$ Match		25 (typ)			30 (typ)		Ω	
$V_{ANALOG}$	Min. Analog Signal Handling Capability		±11			±10		V	
$I_{D(off)}^+$	Switch OFF Leakage	0.1	0.1	20	0.5	0.5	20	nA	$V_D = +10 V$ , $V_S = -10 V$ .
$I_{S(off)}$	Current	0.1	0.1	20	0.5	0.5	20		$V_D = -10V$ , $V_S = +10 V$
$I_{D(on)}^+$	Switch On Leakage	0.2	0.2	40	1	1	40	nA	$V_D = V_S = -10 V$ to +10 V
$I_{S(on)}$	Current								
$t_{on}$	Switch "ON" Time	See switching time specifications and timing diagrams.							
$t_{off}$	Switch "OFF" Time								
$Q_{(INJ.)}$	Charge Injection		100			150		pC	See Fig. 4, Note 2
OIRR	Min. Off Isolation Rejection Ratio		54			50		dB	$f = 1 MHz$ , $R_L = 100\Omega$ , $C_L \leq 5 pF$ See Fig. 5, Note 2
$I^+$	+ Power Supply Quiescent Current	1.0	1.0	10.0	10	10	100	μA	$V^+ = +15 V$ , $V^- = -15 V$ , $V_L = +5 V$ See Fig. 6
$I^-$	- Power Supply Quiescent Current	1.0	1.0	10.0	10	10	100	μA	
$I_L$	+5 V Supply Quiescent Current	1.0	1.0	10.0	10	10	100	μA	
$I_{GND}$	Gnd Supply Quiescent Current	1.0	1.0	10.0	10	10	100	μA	
CCRR	Min. Channel to Channel Cross Coupling Rejection Ratio		54			50		dB	One Channel Off; Any Other Channel Switches See Fig. 7, Note 2

Note: 1. Some channels are turned on by high (1) logic inputs and other channels are turned on by low (0) inputs; however 0.8V to 2.4V describes the min. range for switching properly. Refer to logic diagrams to find logical value of logic input required to produce ON or OFF state.

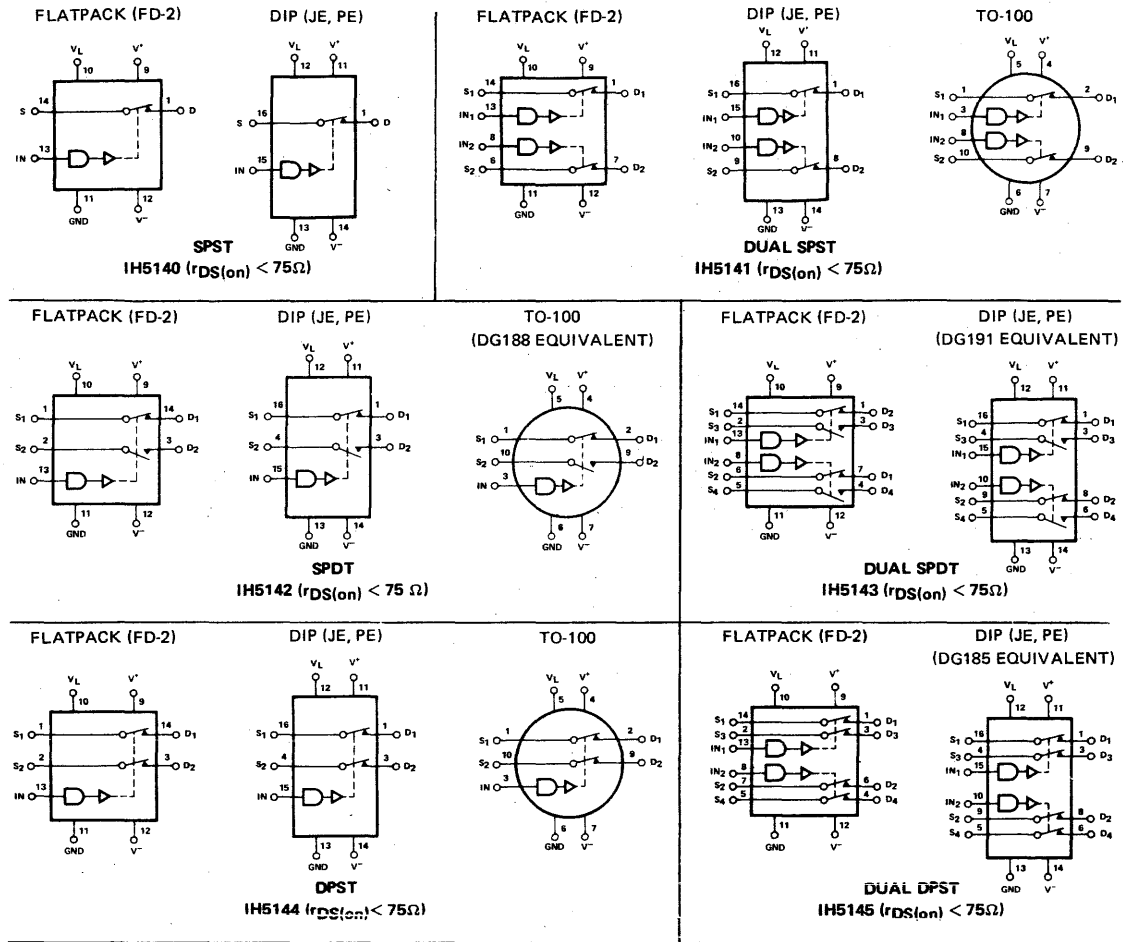
2. Charge injection, OFF isolation, and Channel to Channel isolation are only sample tested in production.



# IH5140-IH5145 Family



**SWITCHING STATE DIAGRAMS** SWITCH STATES ARE FOR LOGIC "1" INPUT





# DG200/IH5200 CMOS Dual SPST Analog Switches

INTERFACE  
Intersil

### FEATURES

- Switches Greater Than 28Vpp Signals With  $\pm 15V$  Supplies
- Break-Before-Make Switching  $t_{off}$  250 nsec,  $t_{on}$  700nsec Typical
- T<sup>2</sup>L, DTL, CMOS, PMOS Compatible
- Non-Latching With Supply Turn-Off
- Complete Monolithic Construction
- Industry Standard (DG200)
- Improved Performance Version (IH5200)

### GENERAL DESCRIPTION

The DG200/IH5200 solid state analog gates are designed using an improved, high voltage CMOS monolithic technology. They provide ease-of-use and performance advantages not previously available from solid state switches. Destructive latch-up of solid state analog gates has been eliminated by INTERSIL's CMOS technology.

The DG200 is completely spec and pin-out compatible with the industry standard device, while the IH5200 offers significantly enhanced specifications with respect to ON and OFF leakage currents, switching times, and supply current.

#### SCHEMATIC DIAGRAM (1/2 DG200/IH5200)

#### PIN CONFIGURATIONS

##### CERDIP & EPOXY DUAL-IN-LINE PACKAGE

(outline dwgs JD, PD)

TOP VIEW

##### METAL CAN PACKAGE

(outline dwg TO-100)

V+ (SUBSTRATE AND CASE)

TOP VIEW

##### FLAT PACKAGE

(outline dwg FD-2)

TOP VIEW

SWITCH STATES ARE FOR LOGIC  
"1" INPUT (POSITIVE LOGIC)

#### ORDERING INFORMATION

INDUSTRY STANDARD PART	IMPROVED SPEC DEVICE	PACKAGE	TEMPERATURE RANGE
DG200AA	IH5200MTW	10-Pin Metal Can	-55 to +125 °C
DG200AK	IH5200MJD	14-Pin CERDIP	-55 to +125 °C
DG200AL	IH5200MFD	14-Pin Flat Pak	-55 to +125 °C
DG200BA	IH5200ITW	10-Pin Metal Can	-25 to +85 °C
DG200BK	IH5200IJD	14-Pin CERDIP	-25 to +85 °C
DG200BL	IH5200IFD	14-Pin Flat Pak	-25 to +85 °C
DG200CJ	IH5200CPD	14-Pin Epoxy DIP	0 to +70 °C



# IH5208

## 4-Channel Differential Fault Protected CMOS Analog Multiplexer

### FEATURES

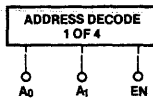
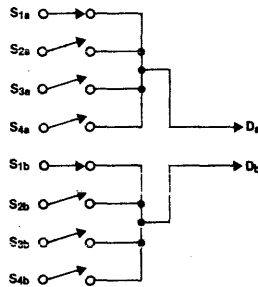
- Ultra low leakage —  $I_{D(off)} \leq 100pA$
- Power supply quiescent current less than  $1\mu A$
- $\pm 13V$  analog signal range
- No SCR latchup
- Break-before-make switching
- TTL and CMOS compatible strobe control
- Pin compatible with HI509, DG509 and AD7509
- All channels OFF ( $I_{ILK} \leq 100nA$ ) when power OFF, for analog signals up to  $\pm 25V$
- Any channel turns OFF ( $I_{ILK} \leq 100nA$ ) if input exceeds supply rails by up to  $\pm 25V$ . Throughput always  $< \pm 14V$  ( $\pm 15V$  supplies)
- TTL and CMOS compatible binary Address and ENable inputs

### GENERAL DESCRIPTION

The IH5208 is a dielectrically isolated CMOS monolithic analog multiplexer, designed as a plug-in replacement for the DG509 and similar devices, but adding fault protection to the standard performance. A unique serial MOSFET switch ensures that an OFF channel will remain OFF when the input exceeds the supply rails by up to  $\pm 25V$ , even with the supply voltage at zero. Further, an ON channel will be limited to a throughput of about 1.5V less than the supply rails, thus affording protection to any following circuitry such as op amps, D/A converters, etc. Cross talk onto "good" channels is also prevented.

A binary 2-bit address code together with the ENable input allows selection of any channel pair or none at all. These 3 inputs are all TTL compatible for easy logic interface; the ENable input also facilitates MUX expansion and cascading.

### FUNCTIONAL DIAGRAM



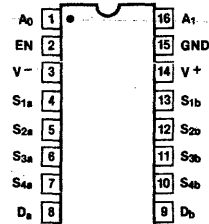
2 LINE BINARY ADDRESS INPUTS  
(0, 0) AND EN = 1  
ABOVE EXAMPLE SHOWS CHANNELS 1a AND 1b ON

### DECODE TRUTH TABLE

A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH PAIR
X	X	0	NONE
0	0	1	1a, 1b
0	1	1	2a, 2b
1	0	1	3a, 3b
1	1	1	4a, 4b

A<sub>0</sub>, A<sub>1</sub>, EN  
Logic "1" = V<sub>AH</sub>  $\geq 2.4V$   
Logic "0" = V<sub>AL</sub>  $\leq 0.8V$

### PIN CONFIGURATION (outline dwg JE, PE)



TOP VIEW

### ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH5208MJE	-55°C to +125°C	16 pin CERDIP
IH5208IJE	-20°C to +85°C	16 pin CERDIP
IH5208CPE	0°C to 70°C	16 pin plastic DIP



# IH5341

## CMOS Monolithic RF/Video Switch

### FEATURES

- $r_{ds(on)} < 75\Omega$ , flat from DC to 100MHz (< 3dB)
- "OFF" isolation > 60dB @ 10MHz
- Cross coupling isolation > 60dB @ 10MHz
- Directly compatible with TTL, CMOS
- Wide operating power supply range
- Power supply current < 1 $\mu$ A
- "Break-before-Make" switching
- Fast switching (80ns/150ns typ)

### GENERAL DESCRIPTION

The IH5341 is a dual SPST, CMOS monolithic switch which uses a "Series/Shunt" ("T" switch) configuration to obtain high "OFF" isolation while maintaining good frequency response in the "ON" condition.

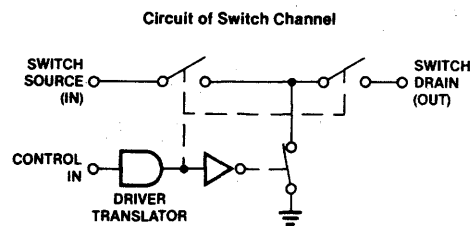
Construction of remote and portable video equipment with extended battery life is facilitated by the extremely low current requirements. Switching speeds are typical  $t_{on} = 150ns$  and  $t_{off} = 80ns$ , and guaranteed "Break-before-Make" switching.

Switch "ON" resistance is typically 40 $\Omega$ -50 $\Omega$  with  $\pm 15V$  power supplies, increasing to typically 175 $\Omega$  for  $\pm 5V$  supplies. The devices are available in TO-100 and 14-pin epoxy DIP packages.

### ORDERING INFORMATION

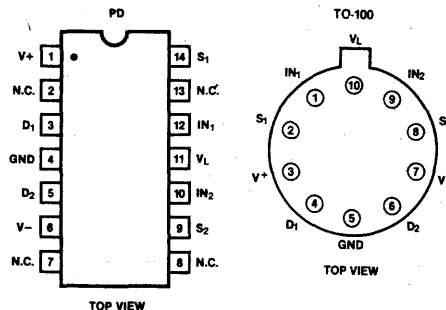
PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH5341CPD	0 to +70°C	14-Pin DIP
IH5341ITW	-20°C to +85°C	TO-100
IH5341MTW	-55°C to +125°C	TO-100

### FUNCTIONAL DIAGRAM



Note: Only one side shown.

### PIN CONFIGURATIONS



**IH5341**



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltages V <sup>+</sup> and V <sup>-</sup> .....	± 17V	Storage Temperature .....	- 65°C to + 160°C
Current in any Terminal .....	50mA	Power Dissipation .....	250mW
Analog Input Voltage .....	V <sup>+</sup> to V <sup>-</sup>	Derate above 25°C @ .....	7.5mW/°C
Operating Temperature		Logic Control Voltage .....	V <sup>+</sup> to V <sup>-</sup>
(M Version) .....	- 55°C to + 125°C	Voltage on Pin 10 .....	V <sup>+</sup> to V <sup>-</sup>
(I Version) .....	- 20°C to + 85°C	Lead Temperature (soldering, 10 seconds) .....	300°C
(C Version) .....	0 to + 70°C		

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**DC ELECTRICAL CHARACTERISTICS** V<sup>+</sup> = + 15V, V<sub>L</sub> = + 5V, V<sup>-</sup> = - 15V, T<sub>A</sub> = 25°C unless otherwise specified.

PARAMETER	SYMBOL	CONDITIONS	TYP	M GRADE DEVICE			I/C GRADE DEVICE			UNITS
				- 55°C	+ 25°C	+ 125°C	- 20/0°C	+ 25°C	+ 85/ + 70°C	
Supply Voltage Ranges										
Positive Supply	V <sup>+</sup>	(Note 3)	4.5 > 16		5 to 15			5 to 15		V
Logic Supply	V <sub>L</sub>		4.5 > V <sup>+</sup>		5 to V <sup>+</sup>		5 to V <sup>+</sup>			
Negative Supply	V <sup>-</sup>		- 4 > - 16		- 5 to - 15		- 5 to - 15			
Switch "ON" Resistance	r <sub>ds(on)</sub>	V <sub>D</sub> = - 5V to + 5V I <sub>S</sub> = 10mA, V <sub>IN</sub> = 2.4V V <sub>D</sub> = - 15V to + 15V		75	75	100	75	75	100	Ω
Switch "ON" Resistance	r <sub>ds(on)</sub>	V <sup>+</sup> = V <sub>L</sub> = 5V, V <sub>IN</sub> = 3V V <sup>-</sup> = 5V, V <sub>D</sub> = ± 5V		250	250	350	300	300	350	
On Resistance Match		I <sub>S</sub> = 10mA, V <sub>D</sub> = ± 5V	5							
Switch "OFF" Leakage	I <sub>D(off)</sub> or I <sub>S(off)</sub>	V <sub>S/D</sub> = + 5V to - 5V V <sub>IN</sub> = 0.8V V <sub>S/D</sub> = + 14V to - 14V		0.1	0.1	20	0.5	0.5	20	nA
Switch "ON" Leakage	I <sub>D(on)</sub> + I <sub>S(on)</sub>	V <sub>D</sub> = + 5V or - 5V V <sub>IN</sub> = 2.4V V <sub>D</sub> = + 14V to - 14V		0.3	0.3	50	1.0	1.0	40	
Input Logic Current	I <sub>IN</sub>	V <sub>IN</sub> > 2.4V or < 0		1	1	10	1	1	10	
Positive Supply Quiescent Current	I <sup>+</sup>	V <sub>IN</sub> = 0V or + 5V		1	1	10	1	1	10	μA
Negative Supply Quiescent Current	I <sup>-</sup>	V <sub>IN</sub> = 0V or + 5V		1	1	10	1	1	10	
Logic Supply Quiescent Current	I <sub>L</sub>	V <sub>IN</sub> = 0V or + 5V		1	1	10	1	1	10	

- Note 1:** Typical values are not tested in production. They are given as a design aid only.
- Note 2:** Positive and negative voltages applied to opposite sides of switch, in both directions successively.
- Note 3:** These are the operating voltages at which the other parameters are tested, and are not directly tested.
- Note 4:** The logic inputs are either greater than or equal to 2.4V or less than or equal to 0.8V, as required, for this test.

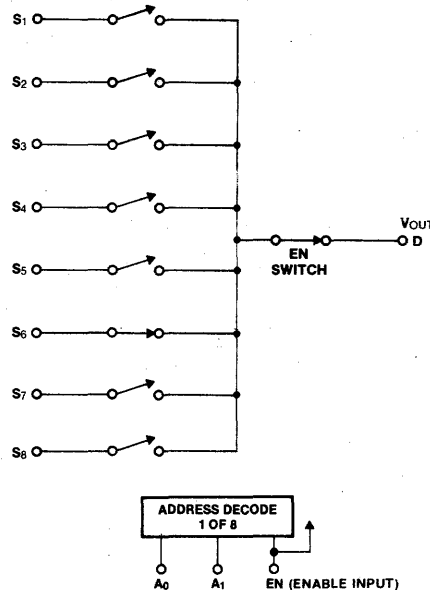
#### FEATURES

- Ultra Low Leakage —  $I_{D(off)} \leq 100\text{pA}$
- $r_{DS(on)} < 400$  ohms over full signal and temperature range
- Power supply quiescent current less than  $100\mu\text{A}$
- $\pm 14\text{V}$  analog signal range
- No SCR latchup
- Break-before-make switching
- Binary Address control (3 Address inputs control 8 channels)
- TTL and CMOS compatible strobe control
- Pin compatible with DG508, HI-508 & AD7508

#### GENERAL DESCRIPTION

The IH6108 is a CMOS monolithic, one of 8 multiplexer. The part is a plug-in replacement for the DG508. Three line binary decoding is used so that the 8 channels can be controlled by 3 Address inputs; additionally a fourth input is provided to use as a system enable. When the ENable input is high (5V) the channels are sequenced by the 3 line Address inputs, and when low (0V) all channels are off. The 3 Address inputs are controlled by TTL logic or CMOS logic elements, a "0" corresponding to any voltage greater than 2.4V. Note that the ENable input (EN) must be taken to 5V to enable the system and less than 0.8V to disable the system.

#### FUNCTIONAL DIAGRAM



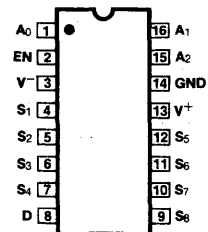
3 LINE BINARY ADDRESS INPUTS  
(1 0 1) AND EN @ 5V  
ABOVE EXAMPLE SHOWS CHANNEL 6 TURNED ON

#### DECODE TRUTH TABLE

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH
x	x	x	0	NONE
0	0	0	1	1
0	0	1	1	2
0	1	0	1	3
0	1	1	1	4
1	0	0	1	5
1	0	1	1	6
1	1	0	1	7
1	1	1	1	8

A<sub>0</sub>, A<sub>1</sub>, A<sub>2</sub>  
Logic "1" = V<sub>AH</sub> ≥ 2.4V    V<sub>ENH</sub> ≥ 4.5V  
Logic "0" = V<sub>AL</sub> ≤ 0.8V

#### PIN CONFIGURATION



#### ORDERING INFORMATION

Ceramic package available as special order only (IH6108MDE/CDE)

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH6108MJE	-55°C to +125°C	16 pin CERDIP
IH6108CJE	0°C to 70°C	16 pin CERDIP
IH6108CPE	0°C to 70°C	16 pin plastic DIP

# IH6108



## ABSOLUTE MAXIMUM RATINGS

V <sub>IN</sub> (A, EN) to Ground	-15V to 15V
V <sub>S</sub> or V <sub>D</sub> to V <sup>+</sup>	0, -32V
V <sub>S</sub> or V <sub>D</sub> to V <sup>-</sup>	0, 32V
V <sup>+</sup> to Ground	16V
V <sup>-</sup> to Ground	-16V
Current (Any Terminal)	30 mA

Current (Analog Source or Drain)	20 mA
Operating Temperature	-55 to 125°C
Storage Temperature	-65 to 150°C
Lead Temp (Soldering, 10 sec)	300°C
Power Dissipation (Package)*	1200 mW

\*All leads soldered or welded to PC board. Derate 10 mW/°C above 70°C.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS V<sup>+</sup> = 15V, V<sup>-</sup> = -15V, V<sub>EN</sub> = +5V (Note 1), Ground = 0V, unless otherwise specified.

CHARACTERISTIC	MEASURED TERMINAL	NO TESTS PER TEMP	TYP 25°C	MAX LIMITS							UNIT	TEST CONDITIONS
				M SUFFIX			C SUFFIX					
				55°C	25°C	125°C	0°C	25°C	70°C			
r <sub>DS(ON)</sub>	S to D	8	180	300	300	400	350	350	450	Ω	V <sub>D</sub> = 10V, I <sub>S</sub> = -1.0mA V <sub>D</sub> = -10V, I <sub>S</sub> = -1.0mA Sequence each switch on V <sub>AL</sub> = 0.8V, V <sub>AH</sub> = 2.4V	
Δr <sub>DS(ON)</sub>			20							%	Δr <sub>DS(ON)</sub> = $\frac{r_{DS(ON)max} - r_{DS(ON)min}}{r_{DS(ON)avg}}$ V <sub>S</sub> = ±10V	
I <sub>S(OFF)</sub>	S	8	0.002		0.05	50		0.1	50	nA	V <sub>S</sub> = 10V, V <sub>D</sub> = -10V V <sub>S</sub> = -10V, V <sub>D</sub> = 10V V <sub>D</sub> = 10V, V <sub>S</sub> = -10V V <sub>D</sub> = -10V, V <sub>S</sub> = 10V V <sub>EN</sub> = 0	
I <sub>D(OFF)</sub>	D	1	0.03		0.1	100		0.2	100	nA	V <sub>S(A1)</sub> = V <sub>D</sub> = 10V V <sub>S(A1)</sub> = V <sub>D</sub> = -10V Sequence each switch on V <sub>AL</sub> = 0.8V, V <sub>AH</sub> = 2.4V	
I <sub>D(ON)</sub>	D	8	0.1		0.2	100		0.4	100	nA	V <sub>A</sub> = 2.4V or 0V V <sub>A</sub> = 15V or 0V	
I <sub>AN(ON)</sub> or I <sub>A(ON)</sub>	A <sub>0</sub> , A <sub>1</sub> or A <sub>2</sub>	3	.01		-10	-30		-10	-30	μA	V <sub>EN</sub> = 5V V <sub>EN</sub> = 0	
I <sub>AN(OFF)</sub> or I <sub>A(OFF)</sub>	Inputs	3	.01		10	30		10	30	μA	All V <sub>A</sub> = 0 (Address pins)	
I <sub>A</sub>	A <sub>0</sub> A <sub>1</sub> A <sub>2</sub>	3			-10	-30		-10	-30	μA	V <sub>EN</sub> = 5V V <sub>EN</sub> = 0	
I <sub>EN</sub>	EN	1			-10	-30		-10	-30	μA	V <sub>EN</sub> = 5V V <sub>EN</sub> = 0	
t <sub>transition</sub>	D		0.3		1						See Fig. 1	
t <sub>lopen</sub>	D		0.2								See Fig. 2	
t <sub>ton(EN)</sub>	D		0.6		1.5						See Fig. 3	
t <sub>toff(EN)</sub>	D		0.4		1							
A "OFF" Isolation	D		60								V <sub>EN</sub> = 0, R <sub>L</sub> = 200Ω, C <sub>L</sub> = 3pF, V <sub>S</sub> = 3 VRMS, f = 500 kHz	
C <sub>S(off)</sub>	S		5								V <sub>S</sub> = 0 V <sub>D</sub> = 0	
C <sub>D(off)</sub>	D		25								V <sub>EN</sub> = 0V, f = 140 kHz to 1 MHz	
C <sub>Ds(off)</sub>	D to S		1								V <sub>S</sub> = 0, V <sub>D</sub> = 0	
Supply Current	+ V <sup>+</sup>	1	40		200			1000			V <sub>EN</sub> = 5V	
Current	- V <sup>-</sup>	1	?		100			1000			V <sub>EN</sub> = 5V	
Standby Current	+ V <sup>+</sup>	1	1		100			1000			V <sub>EN</sub> = 0	
Current	- V <sup>-</sup>	1	1		100			1000			V <sub>EN</sub> = 0	

NOTE 1: See Enable Input Strobing Levels, Section 1.

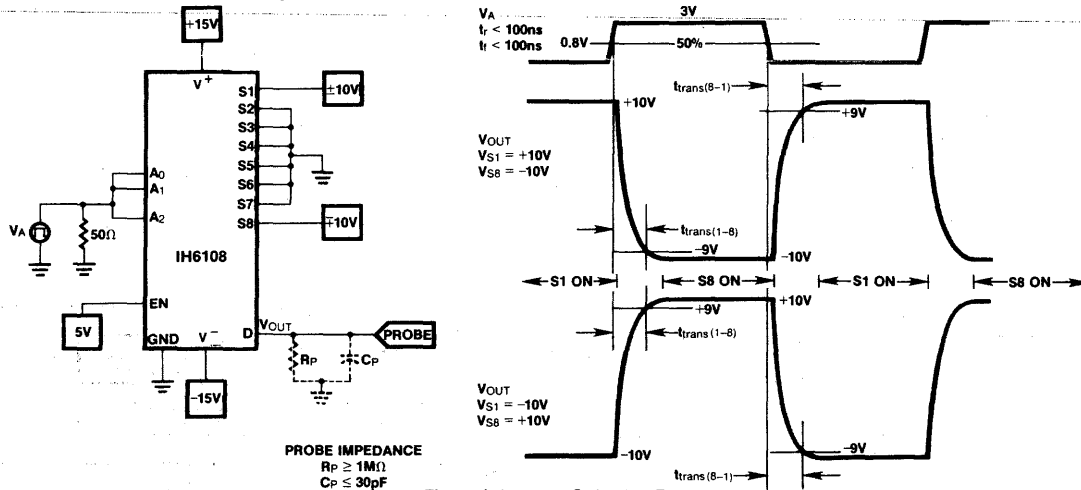


Figure 1. t<sub>transition</sub> Switching Test



# IH6116

## 16-Channel CMOS Analog Multiplexer

INTERFACE

Intersil

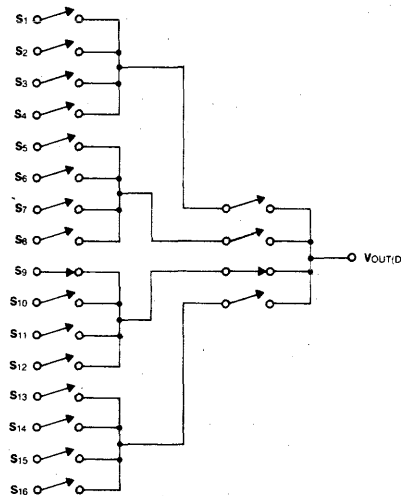
### FEATURES

- Pin compatible with DG506, HI-506 & AD7506
- Ultra Low Leakage —  $I_{D(off)} \leq 100pA$
- $\pm 11$  analog signal range
- $r_{DS(on)} < 700$  ohms over full signal and temperature range
- Break-before-make switching
- TTL and CMOS compatible Address control
- Binary Address control (4 Address inputs control 16 channels)
- Two tier submultiplexing to facilitate expandability
- Power supply quiescent current less than  $100\mu A$
- No SCR latchup

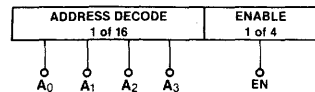
### GENERAL DESCRIPTION

The IH6116 is a CMOS monolithic, one of 16 multiplexer. The part is a plug-in replacement for the DG506. Four line binary decoding is used so that the 16 channels can be controlled by 4 Address inputs; additionally a fifth input is provided to use as system enable. When the ENable input is high (5V) the channels are sequenced by the 4 line Address inputs, and when low (0V), all channels are off. The 4 Address inputs are controlled by TTL logic or CMOS logic elements with a "0" corresponding to any voltage less than 0.8V and a "1" corresponding to any voltage greater than 3.0V. Note that the ENable input must be taken to 5V to enable the system and less than 0.8V to disable the system.

### FUNCTIONAL DIAGRAM



TO DECODE LOGIC CONTROLLING BOTH TIERS OF MUXING



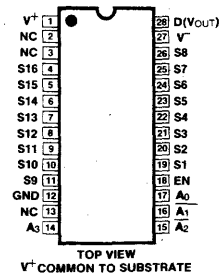
4 LINE BINARY ADDRESS INPUTS  
(0 0 0 1) AND EN @ 5V  
ABOVE EXAMPLE SHOWS CHANNEL 9 TURNED ON

### DECODE TRUTH TABLE

A <sub>3</sub>	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH
x	x	x	x	0	NONE
0	0	0	0	1	1
0	0	0	1	1	2
0	0	1	0	1	3
0	0	1	1	1	4
0	1	0	0	1	5
0	1	0	1	1	6
0	1	1	0	1	7
0	1	1	1	1	8
1	0	0	0	1	9
1	0	0	1	1	10
1	0	1	0	1	11
1	0	1	1	1	12
1	1	0	0	1	13
1	1	0	1	1	14
1	1	1	0	1	15
1	1	1	1	1	16

Logic "1" =  $V_{AH} \geq 3.0V$   $V_{ENH} \geq 4.5V$   
Logic "0" =  $V_{AL} \leq 0.8V$

### PIN CONFIGURATION



TOP VIEW  
V<sup>+</sup> COMMON TO SUBSTRATE

### ORDERING INFORMATION

Ceramic package available as special order only (IH6116MDI/CDI)

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH6116MJI	-55°C to +125°C	28 pin Cerdip
IH6116CJI	0°C to 70°C	28 pin Cerdip
IH6116CPI	0°C to 70°C	28 pin Plastic DIP



# IH6116



## ABSOLUTE MAXIMUM RATINGS

V <sub>IN</sub> (A, EN) to Ground	-15V to 15V
V <sub>S</sub> or V <sub>D</sub> to V <sup>+</sup>	0, -32V
V <sub>S</sub> or V <sub>D</sub> to V <sup>-</sup>	0, 32V
V <sup>+</sup> to Ground	16V
V <sup>-</sup> to Ground	-16V
Current (Any Terminal)	30 mA

Current (Analog Source or Drain)	20 mA
Operating Temperature	-55 to 125°C
Storage Temperature	-65 to 150°C
Lead Temperature (Soldering, 10 secs)	300°C
Power Dissipation (Package)*	1200 mW

\*All leads soldered or welded to PC board. Derate 10 mW/°C above 70°C.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS V<sup>+</sup> = 15V, V<sup>-</sup> = -15V, V<sub>EN</sub> = +5V (Note 1), Ground = 0V, unless otherwise specified.

CHARACTERISTIC	MEASURED TERMINAL	NO TESTS PER TEMP	TYP 25°C	MAX LIMITS						UNIT	TEST CONDITIONS
				M SUFFIX			C SUFFIX				
				-55°C	25°C	125°C	0°C	25°C	70°C		
R <sub>DS(ON)</sub>	S to D	16	480	600	600	700	650	650	750	Ω	V <sub>D</sub> = 10V, I <sub>S</sub> = -10mA V <sub>D</sub> = -10V, I <sub>S</sub> = 10mA Sequence each switch on V <sub>AL</sub> = 0.8V, V <sub>AH</sub> = 3V
		16	300	600	600	700	650	650	750		
ΔR <sub>DS(ON)</sub>			20							%	ΔR <sub>DS(on)</sub> = $\frac{R_{DS(on)max} - R_{DS(on)min}}{R_{DS(on)avg}}$ V <sub>S</sub> = ±10V
I <sub>S(OFF)</sub>	S	16	0.01		0.1	50		0.2	50	nA	V <sub>S</sub> = 10V, V <sub>D</sub> = -10V V <sub>S</sub> = -10V, V <sub>D</sub> = 10V V <sub>D</sub> = 10V, V <sub>S</sub> = -10V V <sub>D</sub> = -10V, V <sub>S</sub> = 10V V <sub>S(AH)</sub> = V <sub>D</sub> = 10V V <sub>S(AL)</sub> = V <sub>D</sub> = -10V Sequence each switch on V <sub>AL</sub> = 0.8V, V <sub>AH</sub> = 3V
		1	0.1		0.2	100		0.4	100		
I <sub>D(OFF)</sub>	D	1	0.1		0.2	100		0.4	100	nA	V <sub>S(AH)</sub> = V <sub>D</sub> = 10V V <sub>S(AL)</sub> = V <sub>D</sub> = -10V Sequence each switch on V <sub>AL</sub> = 0.8V, V <sub>AH</sub> = 3V
		16	0.1		0.2	100		0.4	100		
I <sub>D(ON)</sub>	D	4	.01		-10	-30		-10	-30	μA	V <sub>A</sub> = 3.0V V <sub>A</sub> = 15V
		4	.01		10	30		10	30		
I <sub>A</sub>	A <sub>0</sub> A <sub>1</sub> A <sub>2</sub> A <sub>3</sub>	4			-10	-30		-10	-30	μA	V <sub>EN</sub> = 5V V <sub>EN</sub> = 0 All V <sub>A</sub> = 0
		1			-10	-30		-10	-30		
t <sub>trans</sub>	D		0.6		1					μs	See Fig. 1 See Fig. 2 See Fig. 3
	D		0.2								
	D		0.8		1.5						
	D		0.3		1						
"OFF" Isolation	D		60							dB	V <sub>EN</sub> = 0, R <sub>L</sub> = 200Ω, C <sub>L</sub> = 3pF, V <sub>S</sub> = 3 VRMS, f = 500 kHz
C <sub>S(OFF)</sub>	S		5							pF	V <sub>S</sub> = 0
C <sub>D(OFF)</sub>	D		40							pF	V <sub>D</sub> = 0 V <sub>EN</sub> = 0, f = 140 kHz to 1 MHz
C <sub>ds(OFF)</sub>	D to S		1							pF	V <sub>S</sub> = 0, V <sub>D</sub> = 0
Supply Current	V <sup>+</sup>	1	55		200			1000		μA	V <sub>EN</sub> = 5V V <sub>EN</sub> = 0 A <sub>ii</sub> V <sub>A</sub> = 0 or 3V
	V <sup>-</sup>	1	2		100			1000			
	V <sup>+</sup>	1	1		100			1000			
	V <sup>-</sup>	1	1		100			1000			

NOTE 1: See Section V. Enable Input Strobing Levels.

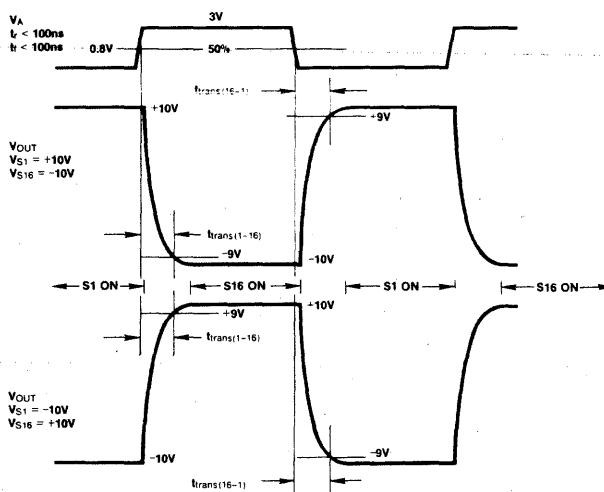
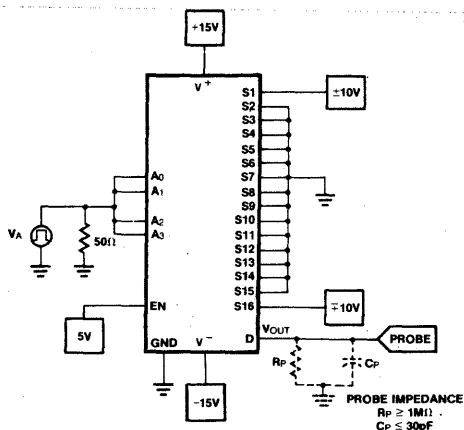


Figure 1

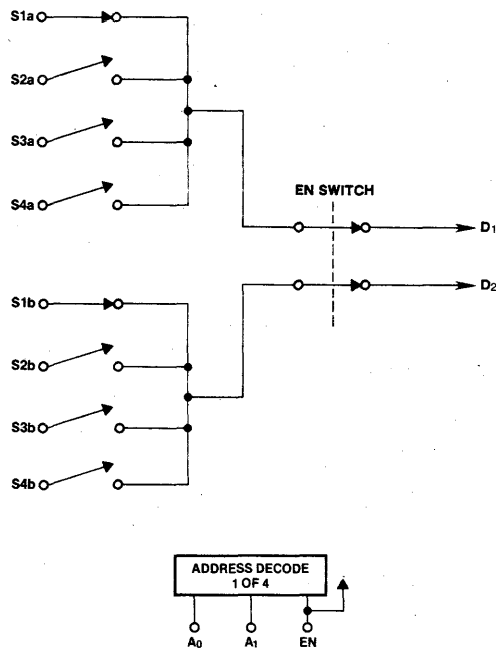
### FEATURES

- Ultra low leakage —  $I_{D(off)} \leq 100\text{pA}$
- $r_{DS(on)} < 400\ \Omega$  over full signal and temperature range
- Power supply quiescent current less than  $100\ \mu\text{A}$
- $\pm 14\text{V}$  analog signal range
- No SCR latch up
- Break-before-make switching
- Binary Address control (2 Address inputs control 2 out of 8 channels)
- TTL and CMOS compatible Address control
- Pin compatible with HI509, DG509 & AD7509

### GENERAL DESCRIPTION

The IH6208 is a monolithic 2 of 8 CMOS multiplexer. The part is a plug-in replacement for the DG509. Two line binary decoding is used so that the 8 channels can be controlled in pairs by the binary inputs; additionally a third input is provided to use as a system enable. When the ENable input is high (5V) the channels are sequenced by the 2 line binary inputs, and when low (0V) all channels are off. The 2 Address inputs are controlled by TTL logic or CMOS logic elements with a "0" corresponding to any voltage less than 0.8V and a "1" corresponding to any voltage greater than 2.4V. Note that the ENable input must be taken to 5V to enable the system, and less than 0.8V to disable the system.

### FUNCTIONAL DIAGRAM



2 LINE BINARY ADDRESS INPUTS  
(0 = 0V, AND EN = 5V (EN = "1" FOR +5V, "0" FOR 0V)  
ABOVE EXAMPLE SHOWS CHANNELS 1a & 1b ON.

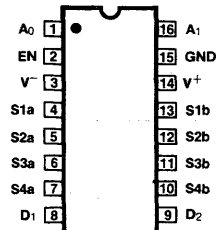
### DECODE TRUTH TABLE

A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH PAIR
X	X	0	NONE
0	0	1	1a, 1b
0	1	1	2a, 2b
1	0	1	3a, 3b
1	1	1	4a, 4b

A<sub>0</sub>, A<sub>1</sub>

LOGIC "1" = V<sub>AH</sub> ≥ 2.4V    V<sub>ENH</sub> ≥ 4.5V  
LOGIC "0" = V<sub>AL</sub> ≤ 0.8V

### PIN CONFIGURATION



### ORDERING INFORMATION

Ceramic package available as special order only (IH6208MDE/CDE)

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH6208MJE	-55°C to +125°C	16 pin CERDIP
IH6208CJE	0°C to 70°C	16 pin CERDIP
IH6208CPE	0°C to 70°C	16 pin Plastic DIP



# IH6216

## 8-Channel Differential CMOS Analog Multiplexer

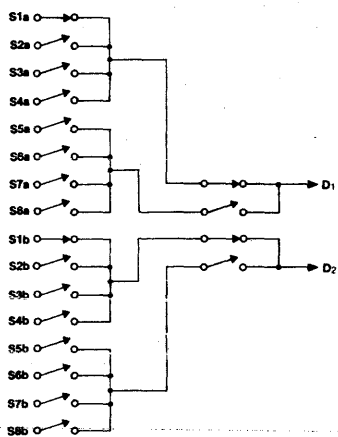
### FEATURES

- Pin compatible with HI507, DG507 & AD7507
- $\pm 11V$  analog signal range
- $r_{DS(on)} < 700$  ohms over full signal and temperature range
- Break-before-make switching
- TTL and CMOS compatible Address control
- Binary Address control (3 Address inputs control 2 out of 16 channels)
- Two tier submultiplexing to facilitate expandability
- Power supply quiescent current less than  $100\mu A$
- No SCR latch up
- Very low leakage  $I_{D(off)} \leq 100pA$

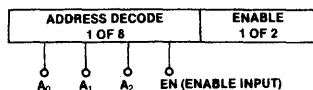
### GENERAL DESCRIPTION

The IH6216 is a CMOS monolithic 2 of 16 multiplexer. The part is a plug-in replacement for the DG507. Three line binary decoding is used so that the 16 channels can be controlled in pairs by the binary inputs; additionally a fourth input is provided to use as a system enable. When the ENable input is high (5V) the channels are sequenced by the 3 line binary inputs, and when low (0V) all channels are off. The 3 Address inputs are controlled by TTL logic or CMOS logic elements with a "0" corresponding to any voltage less than 0.8V and a "1" corresponding to any voltage greater than 3.0V. Note that the ENable input must be taken to 5V to enable the system and less than 0.8V to disable the system.

### FUNCTIONAL DIAGRAM



TO DECODE LOGIC CONTROLLING BOTH TIERS OF MUXING



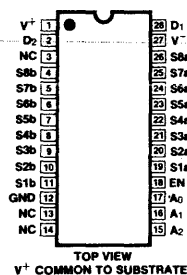
3 LINE BINARY ADDRESS INPUTS (0 0 0) AND EN = 5V  
ABOVE EXAMPLE SHOWS CHANNELS 1a & 1b ON.

### DECODE TRUTH TABLE

A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	ON SWITCH PAIR
X	X	X	0	NONE
0	0	0	1	1
0	0	1	1	2
0	1	0	1	3
0	1	1	1	4
1	0	0	1	5
1	0	1	1	6
1	1	0	1	7
1	1	1	1	8

LOGIC "1" =  $V_{AH} > 3V$   $V_{ENH} > 4.5V$   
LOGIC "0" =  $V_{AL} < 0.8V$

### PIN CONFIGURATION



### ORDERING INFORMATION

Ceramic package available as special order only (IH6216MDI/CDI)

PART NUMBER	TEMPERATURE RANGE	PACKAGE
IH6216MJI	-55°C to +125°C	28 pin Cerdip
IH6216CJI	0°C to 70°C	28 pin Cerdip
IH6216CPI	0°C to 70°C	28 pin Plastic Dip



# IM6402/IM6403 Universal Asynchronous Receiver Transmitter (UART)

## FEATURES

- Low Power — Less Than 10mW Typ. at 2MHz
- Operation Up to 4MHz Clock (IM6402A)
- Programmable Word Length, Stop Bits and Parity
- Automatic Data Formatting and Status Generation
- Compatible with Industry Standard UART's (IM6402)
- On-Chip Oscillator with External Crystal (IM6403)
- Operating Voltage —  
 IM6402-1/03-1: 5V  
 IM6402A/03A: 4-11V  
 IM6402/03: 5V

## GENERAL DESCRIPTION

The IM6402 and IM6403 are CMOS/LSI UART's for interfacing computers or microprocessors to asynchronous serial data channels. The receiver converts serial start, data, parity and stop bits to parallel data verifying proper code transmission, parity, and stop bits. The transmitter converts parallel data into serial form and automatically adds start, parity, and stop bits.

The data word length can be 5, 6, 7 or 8 bits. Parity may be odd or even, and parity checking and generation can be inhibited. The stop bits may be one or two (or one and one-half when transmitting 5 bit code). Serial data format is shown in Figure 6.

The IM6402 and IM6403 can be used in a wide range of applications including modems, printers, peripherals and remote data acquisition systems. CMOS/LSI technology permits clock frequencies up to 4.0MHz (250K Baud), an improvement of 10 to 1 over previous PMOS UART designs. Power requirements, by comparison, are reduced from 670mW to 10mW. Status logic increases flexibility and simplifies the user interface.

The IM6402 differs from the IM6403 in the use of five device pins as indicated in Table 1 and Figure 1.

## PIN CONFIGURATION (outline dwg DL, PL)

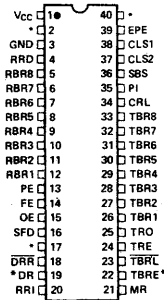


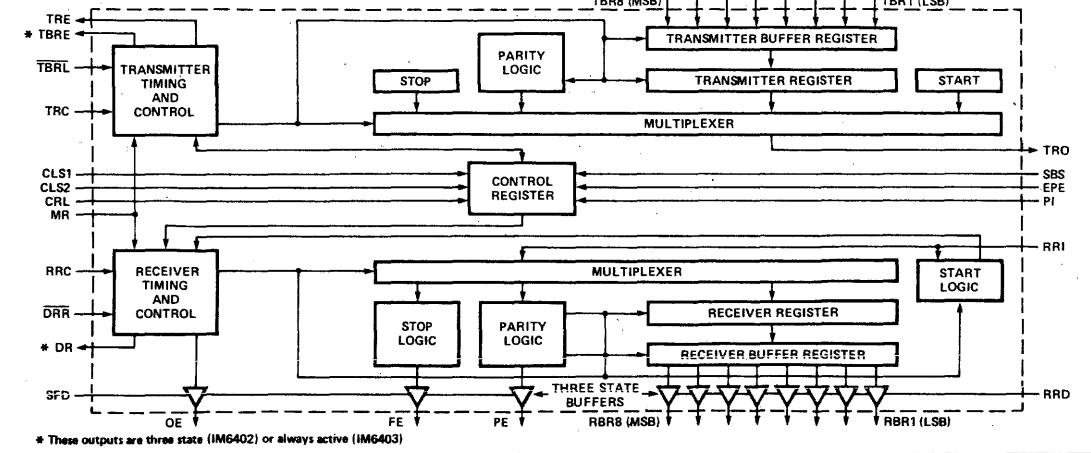
TABLE 1

PIN	IM6402	IM6403 w/XTAL	IM6403 w/EXT CLOCK
2	N/C	Divide Control	Divide Control
17	RRC	XTAL	External Clock Input
19	Tri-State	Always Active	Always Active
22	Tri-State	Always Active	Always Active
40	TRC	XTAL	GND

## ORDERING INFORMATION

ORDER CODE	IM6402-1/03-1	IM6402A/03A	IM6402/03
PLASTIC PKG	IM6402-1/03-1IPL	IM6402/03-AIPL	IM6402/03-IPL
CERAMIC PKG	IM6402-1/03-1IDL	IM6402/03-AIDL	IM6402/03-IDL
MILITARY TEMP.	IM6402-1/03-1MDL	IM6402/03-AMDL	—
MILITARY TEMP. WITH 883B	IM6402-1/03-1MDL/883B	IM6402/03-AMDL/883B	—

## FUNCTIONAL BLOCK DIAGRAM



\* These outputs are three state (IM6402) or always active (IM6403)



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The Micro Networks Division of Unitrode Corporation is an established world leader in the design, development and manufacturing of high-performance, thin and thick-film, hybrid microcircuits for data acquisition/conversion applications. We are the largest supplier of such devices for military/aerospace applications, and are rapidly developing new processing technologies that will enable us to extend our reputation for quality and reliability to the high volume commercial/industrial marketplace. In the first half of 1984, for example, we will introduce the MN565A 12-bit D/A - our first monolithic data converter.

- Analog to Digital Converters
- Digital to Analog Converters
- Track and Hold Amplifiers
- Data Acquisition Systems
- Instrumentation Amplifiers
- Programmable-Gain Amplifiers

**Data Acquisition System Selection Guide**

RESOLUTION	DESCRIPTION	MODEL#	INPUT CHANNELS	ACQUISITION TIME to $\pm 1/2$ LSB ( $\mu$ sec)	CONVERSION TIME ( $\mu$ sec)	THROUGHPUT (Channels/sec)	MAXIMUM LINEARITY ERROR (%FSR)	POWER (mW)	PKG.	SPECIFIED TEMP RANGE ( $^{\circ}$ C)	MIL-STD-883 HI-REL OPTION
8 Bits	Complete 8 bit DAS, 8-Channel Mux, T/H, A/D, 3-State Buffer	MN7120	8 (Expandable)	5	6	90,000	$\pm 0.2$	680	32 Pin	0 to +70 -55 to +125	Yes
12 Bits	10-Channel Multiplexer, Instrumentation Amp and T/H in DIP Package (Use with 12-Bit A/D to Configure a Low-Cost 2 Package DAS)	MN7130	16 (Expandable)	6.5	—	—	$\pm 0.002$	900	32 Pin	0 to +70 -55 to +125	Yes
	Complete 12-Bit DAS, Industry Standard 40-Pin DIP, for Military/Aerospace Applications	MN7140	8 (Expandable)	8	40	20,000	$\pm 0.012$	900	40 Pin	0 to +70 -25 to +85 -55 to +125	Yes
	Complete 8 or 16-Channel DAS, Instru. Amp, T/H, A/D, 3-State Buffers, HDAS-8/16 Compatible	MN7150-8 MN7150-16	8 16 (Expandable)	9	9	50,000	$\pm 0.012$	1875	62 Pin	0 to +70 -25 to +85 -55 to +125	Yes

✓ = New Product

**Instrumentation and Programmable-Gain Amplifier Selection Guide**

GAIN RANGES	DESCRIPTION	MODEL#	MAX GAIN ERROR (%)	OFFSET VOLTAGE (RTI, $\mu$ V)	OFFSET DRIFT ( $\mu$ V/ $^{\circ}$ C)	SMALL SIGNAL BW (kHz)	POWER (mW)	SPECIFIED TEMP RANGE ( $^{\circ}$ C)	PKG.	MIL-STD-883 HI-REL OPTION
1, 10, 100, 1000 Internal, 1 to 1000 with External Resistor	Precision Amp with Internal Gain Scaling Resistors, Excellent Gain Accuracy Over Temp.	MN2200	$\pm 0.01$ to $\pm 0.1$ Depending on Range	$\pm 100$ (G = 100)	$\pm 0.6$ (G = 100)	750 (G = 1)	240	-25 to +85 -55 to +125	18 Pin	Yes
1, 2, 4, 8, 16, 32, 64, 128 Digitally Programmed	Digitally-Controlled Programmable-Gain Amplifier. Binary Weighted Gains Are Set with a 3-Bit Digital Word	MN2020	$\pm 0.005$ to $\pm 0.2$ Depending on Range	$\pm 100$	$\pm 5$	5000 (G = 1)	275	0 to +70 -55 to +125	18 Pin	Yes

# A/D Converter Selection Guide

RESOLUTION	DESCRIPTION	MODEL#	MAXIMUM CONVERSION TIME ( $\mu$ sec) (1)	INTERNAL CLOCK	SPECIFIED TEMP RANGE ( $^{\circ}$ C)	MAXIMUM LINEARITY ERROR (%FSR)	NO MISSING CODES OVER TEMP	INPUT RANGES (Volts)	POWER (mW)	PKG.	MIL-STD-883 HI-REL OPTION
16 Bits	High Performance Guaranteed Over Extended Temperatures	MN5290 MN5291	40	Yes	0 to +70 -25 to +85 -55 to +125	$\pm 0.003$	Yes (2) Yes (3)	6 Ranges	830	32 Pin	Yes
	General Purpose, Fast, Low Cost	MN5280 MN5282	100 50	Yes	0 to +70	$\pm 0.003$	Yes (3)	6 Ranges	1400	32 Pin	No
	Ultra Low Power, 15-Bit Performance, $\pm 12V$ Supplies	MN5284	50	Yes	0 to +70 -25 to +85	$\pm 0.003$	Yes(4)	4 Ranges	150	40 Pin	Yes
14 Bits	Ultra Low Power, TTL and CMOS Compatible, 24-Pin DIP	MN5260	250	No	0 to +70	$\pm 0.003$	Yes (3)	$\pm 10$	215	24 Pin	No
12 Bits	Serial-Parallel Technique, Fastest 12-Bit DIP A/D Available, TTL, 3-State	MN5247 MN5248	0.5	(5)	0 to +70 -25 to +85 -55 to +125	$\pm 0.012$	Yes	0 to +5 $\pm 2.5$	2055	40 Pin	Yes
	Serial-Parallel Technique, Extremely Fast, TTL Compatible	MN5245 MN5246	1	(5)	0 to +70 -25 to +85 -55 to +125	$\pm 0.012$	Yes	0 to +5 $\pm 2.5$	2055	40 Pin	Yes
	Low-Cost, ADC85 Pin-Compatible, Internal or External Clock	MN5243 MN5244	2	Yes No	0 to +70 -25 to +85 -55 to +125	$\pm 0.012$	Yes	6 Ranges	2000	32 Pin	Yes
	Fast, Internal Clock, No Missing Codes, ADC85 Pin-Compatible	MN5240	5	Yes	0 to +70 -25 to +85	$\pm 0.012$	Yes	5 Ranges	1400	32 Pin	Yes
	Fast, Internal Clock, Industry Standard for Industrial Applications	ADC84 ADC85	8	Yes	0 to +70 -25 to +85	$\pm 0.012$	Yes	5 Ranges	1400	32 Pin	Yes
	ADC85 Pin-Compatible, Guaranteed $-55^{\circ}$ C to $+125^{\circ}$ C Operation	ADC87	10	Yes	-25 to +85 -55 to +125	$\pm 0.012$	Yes	5 Ranges	1075	32 Pin	Yes
	Most Widely Used 12-Bit A/D for Military/Aerospace Applications, Fast	MN5210 Series	13	No	0 to +70 -55 to +125	$\pm 0.012$	Yes	4 Ranges	695	24 Pin	Yes
	Industry-Standard Military A/D in Leadless Package	MN5610 Series	13	No	0 to +70 -55 to +125	$\pm 0.012$	Yes	4 Ranges	695	24 Pin Leadless	Yes
	Industry Standard, Low Cost, Complete A/D	ADC80	25	Yes	0 to +70	$\pm 0.012$	Yes	5 Ranges	950	32 Pin	No
	$\mu$ P-Compatible, Address Decoding, 3-State, Low Cost	MN574A	35	Yes	0 to +70 -55 to +125	$\pm 0.012$	Yes	4 Ranges	450	28 Pin	Yes
	Lower Cost Version of Industry Standard MN5210 Series	MN5200 Series	50	No	0 to +70 -55 to +125	$\pm 0.012$	Yes	4 Ranges	695	24 Pin	Yes
	Extremely Low Power, Widely Used in Satellite and Battery Applications	MN5250 Series	175	No	0 to +70 -55 to +125	$\pm 0.012$	Yes	4 Ranges	56	24 Pin	Yes
	Fully Guaranteed Performance $-55^{\circ}$ C to $+200^{\circ}$ C	MN5700	250	No	-55 to +200	$\pm 0.012$	Yes (6)	4 Ranges	311	32 Pin	Yes
8 Bits	Succ. Approx. A/D High Speed, Low Cost	MN5815 MN5825	0.7 1	Yes	0 to +70 -55 to +125	$\pm 0.2$	Yes	6 Ranges	925	24 Pin	Yes
	Very Fast, Widely Used in Avionics Applications	MN5101 MN5100	0.9 1.5	No	0 to +70 -55 to +125	$\pm 0.2$	Yes	9 Ranges	1125	24 Pin	Yes
	Fast, Low Cost Popular, 18-Pin DIP	MN5130 Series	2.5	No	0 to +70 -55 to +125	$\pm 0.2$	Yes	4 Ranges	680	18 Pin	Yes
	$\pm 12V$ Supplies for Microprocessor and CMOS Applications	MN5140 Series	2.5	No	0 to +70 -55 to +125	$\pm 0.2$	Yes	4 Ranges	680	18 Pin	Yes
	Internal 3-State Output Buffer for Data Bus Applications	MN5150	2.5	No	0 to +70 -55 to +125	$\pm 0.2$	Yes	7 Ranges	680	24 Pin	Yes
	Most Popular 8-Bit A/D for Industrial and Military Applications	MN5120 Series	6	No	0 to +70 -55 to +125	$\pm 0.2$	Yes	4 Ranges	680	18 Pin	Yes
	Ultra Low Power, Single $+12V$ Supply, TTL and CMOS Compatible	MN5065 MN5066	100	No	0 to +70 -55 to +125	$\pm 0.2$	Yes	$\pm 5$ 0 to +10	53	18 Pin	Yes

✓ = New Product

- For units with an external clock, the table shows the minimum conversion time that will result in specified accuracy. For units with internal clock, maximum conversion time is given.
- No Missing Codes for 14 bits guaranteed over temperature.
- No Missing Codes for 14 bits guaranteed at  $+25^{\circ}$ C.  
No Missing Codes for 13 bits guaranteed over temperature.
- No Missing Codes for 15 bits guaranteed over temperature.
- The serial-parallel conversion technique does not require a clock.
- No Missing Codes for 10 bits guaranteed at  $+200^{\circ}$ C.



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# D/A Converter Selection Guide

RESOLUTION	DESCRIPTION	MODEL#	MAXIMUM SETTLING TIME ( $\mu$ sec) (1)	INTERNAL REF. AND OUTPUT OP AMP	SPECIFIED TEMP RANGE ( $^{\circ}$ C)	MAXIMUM LINEARITY ERROR (%FSR)	MONOTONIC OVER TEMP	OUTPUT RANGES (Volts)	POWER (mW)	PKG.	MIL-STD-883 HI-REL OPTION
16 Bits	Industry Standard for Industrial Applications	DAC71	10	Yes (2)	0 to +70	$\pm 0.003$	(Note 3)	2V Ranges 2I Ranges	850	24 Pin	No
12 Bits	Industry-Standard Monolithic D/A, Current Output, On-Chip Reference	MN565A	0.25	Yes (2)	0 to +70 -55 to +125	$\pm 0.012$	Yes	0 to -2mA $\pm 1mA$	225	24 Pin	Yes
	Industry Standard, Low Cost	DAC80	3 (Typ)	Yes	0 to +70	$\pm 0.012$	Yes	5V Ranges 2I Ranges	850	24 Pin	No
	Industry Standard, Binary Coding, Voltage/Current Output	DAC85-CBI	3 (Typ)	Yes	0 to +70 -25 to +85	$\pm 0.012$	Yes	5V Ranges 2I Ranges	850	24 Pin	Yes
	Industry Standard, 3 Decade BCD, Voltage/Current Output	DAC85-CCD	3 (Typ)	Yes	0 to +70 -25 to +85	$\pm 0.012$	Yes	0 to +9.99 0 to -1.25mA	850	24 Pin	Yes
	Industry-Standard 12-Bit D/A for Military/Aerospace Applications	DAC87	4	Yes	-25 to +85 -55 to +125	$\pm 0.012$	Yes	5 Ranges	900	24 Pin	Yes
	12-Bit D/A With Fast Input Register	DAC HK (4)	4 (Typ)	Yes	0 to +70 -55 to +125	$\pm 0.0012$	Yes	5 Ranges	975	24 Pin	Yes
	Military-Standard Latched D/A in Leadless Package	MN3660 (5)	10	Yes	0 to +70 -55 to +125	$\pm 0.012$	Yes	0 to +10 $\pm 5, \pm 10$	495	24 Pin Leadless	Yes
	For Military/Aerospace and Other High Reliability Applications	MN3850	7	Yes	0 to +70 -55 to +125	$\pm 0.012$	Yes	5 Ranges	525	24 Pin	Yes
	Fast Input Register, For Military/Aerospace Hi-Rel Applications	MN3860 (5)	7	Yes	0 to +70 -55 to +125	$\pm 0.012$	Yes	0 to +10 $\pm 5, \pm 10$	675	24 Pin	Yes
	Low-Cost, Low-Power, Superior Second Source for the DAC349	MN3349	10	Yes	0 to +70 -55 to +125	$\pm 0.012$	(Note 6)	5 Ranges	195	24 Pin	Yes
	Low Cost, Low Power, DAC349 Compatible	MN3348	8	Yes	0 to +70 -55 to +125	$\pm 0.012$	(Note 6)	5 Ranges	195	24 Pin	Yes
	12-Bit D/A Converter with Fast Input Register, Hi-Rel and Industrial Applications	DAC88 (5)	10	Yes	0 to +70 -55 to +125	$\pm 0.012$	Yes	0 to +10 $\pm 5, \pm 10$	495	24 Pin	Yes
	4 Quadrant Multiplying, Voltage Output, High Accuracy	MN3412	20	(Note 7)	0 to +70 -55 to +125	$\pm 0.012$	Yes	$\pm 10$	335	32 Pin Module	Yes
	Industry-Standard Low-Power D/A, Voltage Output	MN370 MN371	60 35	Yes	0 to +70 -55 to +125	$\pm 0.012$	Yes	-10 to +10 0 to +10	60	18 Pin	Yes
10 Bits	Fast Input Latch for Hi-Rel $\mu$ P Applications	MN3040 (5)	10	Yes	0 to +70 -55 to +125	$\pm 0.05$	Yes	0 to -10 $\pm 10$	450	18 Pin	Yes
	Excellent for Servo Applications, Very Accurate Zero	MN3003 Series	30	Yes	0 to +70 -55 to +125	$\pm 0.05$	Yes	4 Ranges	450	16 Pin	Yes
8 Bits	Produces Composite Video Signal for Raster-Scan Displays	MN0805	8nsec	Yes (2)	-25 to +85	$\pm 0.2$	Yes	0 to -17mA (8)	1660	24 Pin	Yes
	Fast Settling, 30 V/ $\mu$ sec Slew Rate	MN3008 MN3009	1	Yes	0 to +70 -55 to +125	$\pm 0.2$	Yes	0 to +4 $\pm 2$	495	16 Pin	Yes
	Fast Input Latch, Popular in Avionics Applications, Complete in 18-Pin DIP	MN3020 (5)	3	Yes	0 to +70 -55 to +125	$\pm 0.2$	Yes	4 Ranges	505	18 Pin	Yes
	Popular, Low Cost, General Purpose, Multirange	MN3013 MN3014	30 2.5	Yes	0 to +70 -55 to +125	$\pm 0.2$	Yes	4 Ranges	420	16 Pin	Yes
	Excellent for Servo Applications, Very Accurate Zero, Small 14-Pin DIP	MN3000 Series	30	Yes	0 to +70 -55 to +125	$\pm 0.2$	Yes	4 Ranges	450	14 Pin	Yes
6 Bits	Composite Video for Raster-Scan Displays	MN0605	6nsec	Yes (2)	-25 to +85	$\pm 0.8$	Yes	0 to -17mA (8)	1350	24 Pin	Yes
4 Bits	Composite Video for Raster-Scan Displays	MN0405	4nsec	Yes (2)	-25 to +85	$\pm 3.2$	Yes	0 to -17mA (8)	1040	24 Pin	Yes

✓ = New Product

1. Full scale change settling to  $\pm 1/2$ LSB.
2. Current-output devices have no output op amps
3. Monotonicity for 14 bits guaranteed over temperature.
4. Latch timing parameters: minimum setup time 50nsec, maximum hold time 0nsec, minimum latch enable pulse width 60nsec.
5. Latch timing parameters: minimum setup time 40nsec, maximum hold time 0nsec, minimum latch enable pulse width 60 nsec.
6. Monotonicity guaranteed at 25 $^{\circ}$ C.
7. Internal output op amp.
8. Produces standard 1V-p-p video signal when driving 75 $\Omega$  load.



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# Track-Hold Amplifier Selection Guide

## 7-9 Bit Applications

MAXIMUM LINEARITY ERROR (%)	ACQUISITION TIME	GAIN AND VOLTAGE RANGE	DESCRIPTION	MODEL #	SPECIFIED TEMP RANGE (°C)	APERTURE JITTER (psec, max)	DROOP RATE ( $\mu\text{V}/\mu\text{s}$ )	POWER (mW)	PKG.	MIL-STD-883 HI-REL OPTION
± 0.1	35nsec 5V Step to ± 0.1%	+1, ± 2.5V	Ultra Fast, Designed to Drive Flash Converters, 500pF Cap Loading	MN379	0 to +70 -55 to +125	2	± 500	1575	24 Pin	Yes

## 10-12 Bit Applications

MAXIMUM LINEARITY ERROR (%)	ACQUISITION TIME	GAIN AND VOLTAGE RANGE	DESCRIPTION	MODEL #	SPECIFIED TEMP RANGE (°C)	APERTURE JITTER (psec, max)	DROOP RATE ( $\mu\text{V}/\mu\text{s}$ )	POWER (mW)	PKG.	MIL-STD-883 HI-REL OPTION
± 0.01	100nsec 5V Step to ± 0.1%	-1, ± 5V	Package Contains User-Optional Pin-Programmable Input Buffer	MN050	0 to +70 -55 to +125	100	± 0.1	1400	24 Pin	Yes

## 12-Bit Applications

MAXIMUM LINEARITY ERROR (%)	ACQUISITION TIME	GAIN AND VOLTAGE RANGE	DESCRIPTION	MODEL #	SPECIFIED TEMP RANGE (°C)	APERTURE JITTER (psec, max)	DROOP RATE ( $\mu\text{V}/\mu\text{s}$ )	POWER (mW)	PKG.	MIL-STD-883 HI-REL OPTION
± 0.01	90nsec 10V Step to ± 0.01%	-1, ± 10V	Fastest ± 0.01% T/H Available, TTL Compatible	MN377	0 to +70 -55 to +125	50	± 2	875	24 Pin	Yes
± 0.01	160nsec 10V Step to ± 0.01%	-1, ± 10V	Second Fastest ± 0.01% T/H Available, TTL Compatible	MN376	0 to +70 -55 to +125	100	± 0.5	730	24 Pin	Yes
± 0.01	250nsec 10V Step to ± 0.01%	-1, ± 10V	Third Fastest ± 0.01% T/H, HTC-0300A Compatible	MN0300A	0 to +70 -55 to +125	100	± 5	730	24 Pin	Yes
± 0.005	500nsec 10V Step to ± 0.01%	-1, ± 5V	Precision T/H Suitable for High-Speed 12-14 Bit Applications	MN375	0 to +70 -25 to +85	100	± 0.5	1325	24 Pin	Yes
± 0.01	1 $\mu$ sec 10V Step to ± 0.01%	-1, ± 10V	Fast, Low Cost, Internal Hold Capacitor, 14-Pin DIP	MN346	0 to +70 -55 to +125	30 (1)	± 0.1	640	14 Pin	Yes
± 0.01	1 $\mu$ sec 10V Step to ± 0.05%	-1, ± 10V	Fast, Low Cost, Internal Hold Capacitor, 14-Pin DIP	MN347	0 to +70 -55 to +125	30 (1)	± 0.5	640	14 Pin	Yes
± 0.01	6.5 $\mu$ sec 20V Step to ± 0.01%	-1, ± 10V	16-Channel Mux, Instru. Amp and T/H in a Single DIP (Use with A/D to Make 2 Package DAS)	MN7130	0 to +70 -55 to +125	60 (1)	± 4	900	32 Pin	Yes
± 0.01	7.5 $\mu$ sec 10V Step to ± 0.01%	-1, ± 10V	Moderate Speed, Very Low Droop, Internal Hold Capacitor	MN343	0 to +70 -55 to +125	60 (1)	± 0.1	345	14 Pin	Yes
± 0.01	7.5 $\mu$ sec 10V Step to ± 0.05%	-1, ± 10V	Low Cost, Low Droop Internal Hold Capacitor	MN344	0 to +70 -55 to +125	60 (1)	± 0.4	345	14 Pin	Yes

(1) Listed Specification is Aperture Delay Time

## 13-16 Bit Applications

MAXIMUM LINEARITY ERROR (%)	ACQUISITION TIME	GAIN AND VOLTAGE RANGE	DESCRIPTION	MODEL #	SPECIFIED TEMP RANGE (°C)	APERTURE JITTER (psec, max)	DROOP RATE ( $\mu\text{V}/\mu\text{s}$ )	POWER (mW)	PKG.	MIL-STD-883 HI-REL OPTION
± 0.005	700nsec 10V Step to ± 0.005%	-1, ± 5V	Precision T/H Suitable for 12-14 Bit Applications	MN375	0 to +70 -25 to +85	100	± 0.5	1325	24 Pin	Yes
± 0.003	8.5 $\mu$ sec 10V Step to ± 0.003%	-1, ± 10V	Precision T/H Compatible with MN5290, MN5291, MN5282 and other 16-Bit DIP A/D's	MN373	0 to +70 -55 to +125	1000	± 0.5	300	14 Pin	Yes

✓ = New Product



## Micro Networks

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Note: See other MPS parts in Linear and Military Sections.

## Digital-to-Analog Converters

- Second source to Analog Devices and PMI
- Improved temperature drift characteristics
- Superior stability — no laser trimming on monolithic DAC's
- .5 ppm gain tempco for MP7621

Model	Type	Resolution	Accuracy	Gain Tempco	Settling Time (typ)	Output	Power Dissipation	Replaces
MP370	CMOS, High Resolution	18 Bit	±.0008%	2 ppm	2 μsec	Current	60 mW (typ)	DAC-370-18
MP377-18	CMOS, Buffered, Complete	18 Bit	±.0008%	5 ppm	20 μsec	Voltage	500 mW (max)	DAC-377-18
MP562	Bipolar, Low Power	12 Bit	±.006%	5 ppm	1.5 μsec	Current	200 mW (typ)	AD562
MP1208	CMOS, μP Compatible	12 Bit	±.012%	2 ppm	1 μsec	Current	20 mW (typ)	DAC1208
MP1230	CMOS, μP Compatible	12 Bit	±.012%	2 ppm	1 μsec	Current	20 mW (typ)	DAC1230
MP3140	CMOS, Multiplying	14 Bit	±.006%	2 ppm	2 μsec	Current	30 mW (max)	HS3140
MP5520	CMOS and Bipolar, Internal amp and reference	6 Bit	±.2%	±80 ppm	3 μsec	Voltage	250 mW (max)	DAC01
MP7520	CMOS, Multiplying	10 Bit	±.05%	2 ppm	500 nsec	Current	20 mW (typ)	AD7520
MP7521	CMOS, Multiplying	12 Bit	±.05%	2 ppm	500 nsec	Current	20 mW (typ)	AD7521
MP7522	CMOS, μP Compatible, Double buffered input	10 Bit	±.05%	10 ppm	500 nsec	Current	20 mW (typ)	AD7522
MP7523	CMOS, Multiplying, Low Cost	8 Bit	±.05%	10 ppm	150 nsec	Current	10 mW (typ)	AD7523
MP7524	CMOS, μP Compatible, Latched input	8 Bit	±.05%	10 ppm	100 nsec	Current	10 mW (typ)	AD7524
MP7528	CMOS, Buffered Multiplying	8 Bit	±.02%	10 ppm	80 nsec	Current	15 mW (max)	AD7528
MP7530	CMOS, Multiplying, Low Cost	10 Bit	±.05%	2 ppm	500 nsec	Current	20 mW (typ)	AD7530
MP7531	CMOS, Multiplying, Low Cost	12 Bit	±.05%	2 ppm	500 nsec	Current	20 mW (typ)	AD7531
MP7533	CMOS, Multiplying, Low Cost	10 Bit	±.05%	2 ppm	600 nsec	Current	30 mW (typ)	AD7533
MP7541	CMOS, Multiplying	12 Bit	±.012%	2 ppm	1 μsec	Current	20 mW (typ)	AD7541
MP7542	CMOS, μP Compatible	12 Bit	±.012%	2 ppm	1 μsec	Current	20 mW (typ)	AD7542
MP7543	CMOS, Serial Input	12 Bit	±.012%	2 ppm	1 μsec	Current	20 mW (typ)	AD7543
MP7645	CMOS, Buffered Multiplying	12 Bit	±.012%	2 ppm	1 μsec	Current	20 mW (typ)	AD7545
MP7614	CMOS, Multiplying	14 Bit	±.006%	10 ppm	2 μsec	Current	40 mW (typ)	
MP7616	CMOS, Multiplying	16 Bit	±.006%	10 ppm	2 μsec	Current	40 mW (typ)	
MP7621	CMOS, Untrimmed, Multiplying	12 Bit	±.012%	.5 ppm	1 μsec	Current	20 mW (typ)	AD7541
MP7622	CMOS, Multiplying, Double Buffered	12 Bit	±.012%	2 ppm	1 μsec	Current	40 mW (typ)	HS3120
MP7623	CMOS, Multiplying	12 Bit	±.012%	.5 ppm	1 μsec	Current	20 mW (typ)	AD7541A
MP9331	CMOS, High Resolution	16 Bit	±.0008%	2 ppm	2 μsec	Current	60 mW (typ)	DAC-9331-18
MP9377-16	CMOS, Buffered, Complete	16 Bit	±.0008%	5 ppm	20 μsec	Voltage	500 mW (max)	DAC-9377-16
MP8526	Complete Bipolar DAC	13 Bit	±.006%	10 ppm	5 μsec	Voltage	300 mW	

## Analog-to-Digital Converters

- Second source to Analog Devices
- Bright display digital panel meter
- Microprocessor compatible

Model	Type	Resolution	Accuracy	Conversion Time	Outputs	Power Dissipation	Replaces
MP574	CMOS, Complete, μP Interface	12 Bit	±.012	25 μsec	Parallel, Tristate	600 mW	AD574
MP7138	CMOS 3½ Digit Panel Meter	100 μV	±.05%	100 msec	BCD	10 mW (typ)	—
MP7550	CMOS, Quadslope, 2's Complement	13 Bit	±1 LSB	40 msec	Parallel	9 mW	AD7550
MP7570	CMOS, Successive Approximation	10 Bit	±.05%	40 μsec	Tristate, Serial & Parallel	40 mW	AD7570
MP7574	CMOS, Successive Approximation	8 Bit	±3/4 LSB	15 μsec	Tristate, Binary	25 mW	AD7574
MP7581	CMOS, 8 Channel, 8 Bit	8 Bit	±1/2 LSB		Tristate, Binary	40 mW (max)	AD7581

All Micro Power Systems' IC's are available with 883B processing, in dice form, or to customer's specifications.

Note: See other MPS parts in Linear and Military Sections.

## Analog Multiplexers

- Second source to Harris and Siliconix
- Low leakage; dielectric isolation
- 25 volt over-voltage protection
- Low ON resistance

NOTE: Dielectrically isolated switches are indicated by "DI" suffix.

Model	Type	Temperature Ranges	Packages	RON (max, full temp)	Off Output Leakage (max, full temp)	Access Time (typ, 25°C)	Replaces
MP7501	8 Channel	All	All	375 ohm	250 nA	800 nsec	AD7501
MP7502	4 Channel Diff.	All	All	375 ohm	125 nA	800 nsec	AD7502
MP7503	8 Channel (Inv. Enable)	All	All	375 ohm	250 nA	800 nsec	AD7503
MP7506	16 Channel	All	All	500 ohm	500 nA	800 nsec	AD7506 DG506
MP7507	8 Channel Diff.	All	All	500 ohm	250 nA	800 nsec	AD7507 DG507
MP7508DI	8 Channel	All	All	350 ohm	500 nA	500 nsec	H1508A DG508
MP7509DI	4 Channel Diff.	All	All	350 ohm	500 nA	500 nsec	H1509A DG509

## Analog Switches

- Second source to Analog Devices, Harris and Siliconix
- Low ON resistance

NOTE: Dielectrically isolated switches are indicated by "DI" suffix.

Model	Type	Temperature Ranges	Packages	Off Output Leakage (max, full temp)	RON (max, full temp)	Switch ON Time (typ, 25°C)	Power Dissipation (typ, 25°C)	Replaces
MP200DI	Dual, SPST	All	All	500 nA	100 ohm	240 nsec	15 mW	DG200 H1200 MP7513
MP201DI	Quad, SPST	All	All	500 nA	125 ohm	185 nsec	15 mW	DG201 H1201 MP7516
MP7510DI	Quad, SPST	All	All	500 nA	175 ohm	180 nsec	3 mW	AD7510DI
MP7511DI	Quad, SPST	All	All	500 nA	175 ohm	180 nsec	3 mW	AD7511DI
MP7512DI	Dual, SPDT	All	All	500 nA	175 ohm	180 nsec	3 mW	AD7512DI

All Micro Power Systems' IC's are available with 883B processing, in dice form, or to customer's specifications.



# MD54/74HCT138R

## 1 of 8 Octal Decoder/Demultiplexer

### Preliminary Information

October '83

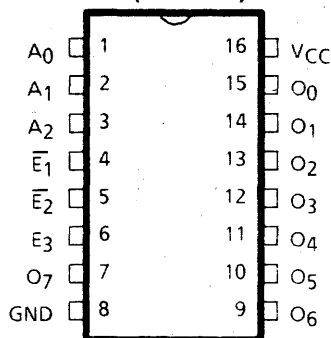
### Features

- High latch-up immunity
- High current outputs can drive 15 LSTTL loads
- Low power ISO-CMOS technology
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS138 types

### Description

This ISO-CMOS, MSI circuit is designed for use in high speed memory and peripheral address decoding systems. The MD54/74HCT138 will decode 3 binary inputs ( $A_0, A_1, A_2$ ) to select one of eight mutually exclusive outputs ( $O_0 - O_7$ ). Three enable inputs, two active LOW ( $\bar{E}_1, \bar{E}_2$ ) and one active HIGH ( $E_3$ ), reduce the need for external gates in an expanded system. A 1 of 32 decoder requires only four of these devices and one external inverter. The enable inputs can be used as the data input for demultiplexing applications.

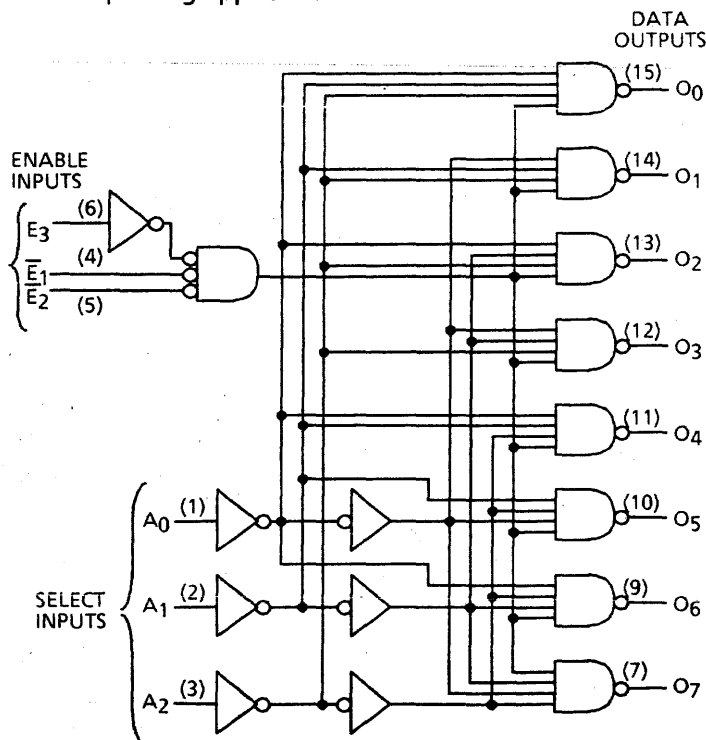
CONNECTION DIAGRAM  
DIP (TOP VIEW)



### Ordering Information

MD54HCT138RCB, Cerdip  
-55°C to 125°C  
with MIL 883B option

MD74HCT138RE, Plastic Dip  
-40°C to 85°C



### FUNCTION TABLE

INPUTS			OUTPUTS								
ENABLE	SELECT										
$E_3 \bar{E}_1 \bar{E}_2$	$A_2$	$A_1$	$A_0$	$O_0$	$O_1$	$O_2$	$O_3$	$O_4$	$O_5$	$O_6$	$O_7$
X H X	X	X	X	H	H	H	H	H	H	H	H
X X H	X	X	X	H	H	H	H	H	H	H	H
L X X	X	X	X	H	H	H	H	H	H	H	H
H L L	L	L	L	L	H	H	H	H	H	H	H
H L L	L	L	H	H	L	H	H	H	H	H	H
H L L	L	H	L	H	H	L	H	H	H	H	H
H L L	L	H	H	H	H	L	H	H	H	H	H
H L L	H	L	L	H	H	H	H	L	H	H	H
H L L	H	L	H	H	H	H	H	L	H	H	H
H L L	H	H	L	H	H	H	H	H	L	H	H
H L L	H	H	H	H	H	H	H	H	H	L	H

H = logic "1", L = logic "0", X = don't care



# MD54/74HCT139R

## Dual 1 of 4 Decoder/Demultiplexer

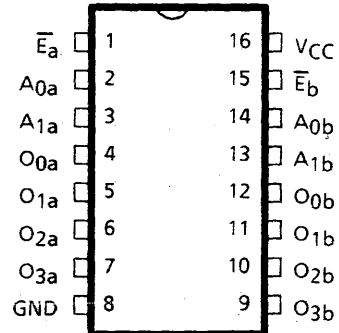
Preliminary Information

October '83

### Features

- High latch-up immunity
- High current outputs can drive 15 LSTTL loads
- Low power ISO-CMOS technology
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS139 types

CONNECTION DIAGRAM  
DIP (TOP VIEW)



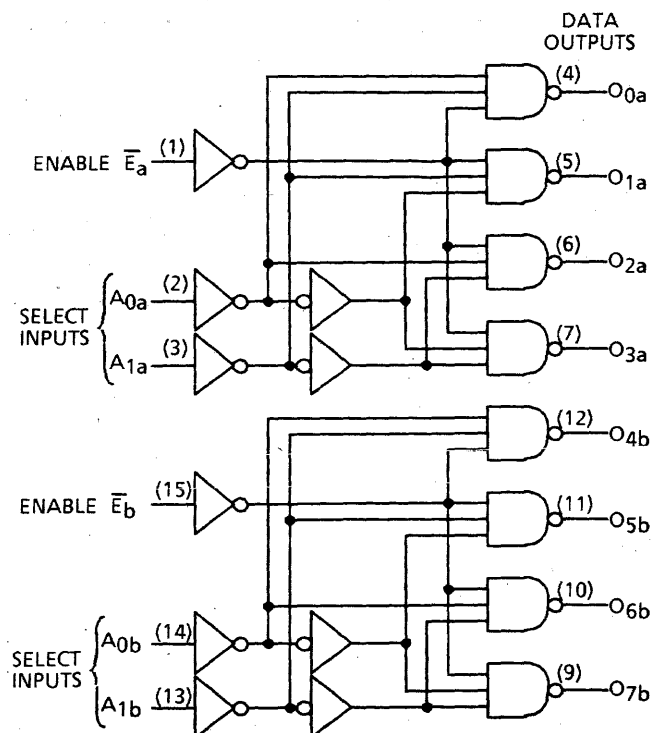
### Description

This ISO-CMOS, MSI circuit is designed for use in high speed memory and peripheral, address decoding systems. The MD54/74HCT139 will decode 2 binary inputs ( $A_0$ ,  $A_1$ ) to select one of four mutually exclusive outputs ( $O_0$  -  $O_3$ ). The enable inputs, active LOW ( $\bar{E}_a$ ,  $\bar{E}_b$ ) independently control the two separate 1 of 4 decoder/demultiplexers.

### Ordering Information

MD54HCT139RCB, Cerdip  
-55°C to 125°C  
with MIL 883B option

MD74HCT139RE, Plastic Dip  
-40°C to 85°C



FUNCTION TABLE  
(each decoder/demultiplexer)

INPUTS		OUTPUTS			
ENABLE	SELECT				
$\bar{E}$	$A_1$ $A_0$	$O_0$	$O_1$	$O_2$	$O_3$
H	X X	H	H	H	H
L	L L	L	H	H	H
L	L H	H	L	H	H
L	H L	H	H	L	H
L	H H	H	H	H	L

H = logic "1", L = logic "0", X = don't care



# MD54/74HCT240R

## Inverting Octal Line Driver/Buffer

### Preliminary Information

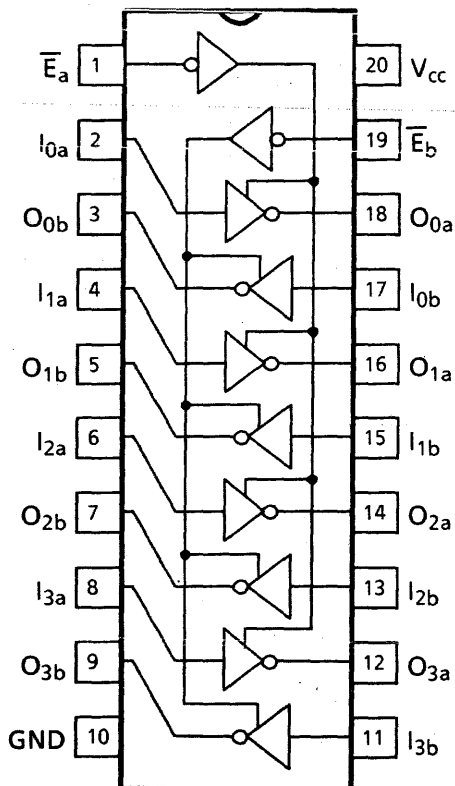
October '83

### Features

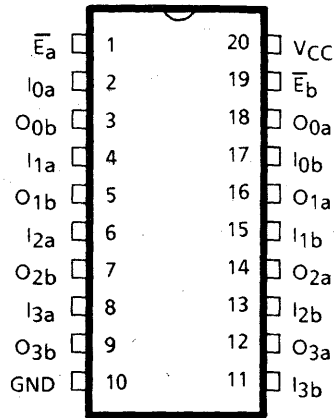
- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS240 types

### Description

These ISO-CMOS Octal line driver/buffers are designed to improve P.C. board density and performance in three-state memory address drivers, clock drivers, and bus oriented receivers/transmitters. The MD54/74HCT240R offers inverting three-state buffers controlled by two active low enable pins.



CONNECTION DIAGRAM  
DIP (TOP VIEW)



### Ordering Information

MD54HCT240RCB, Cerdip  
-55°C to 125°C  
with MIL 883B option

MD74HCT240RE, Plastic Dip  
-40°C to 85°C

PIN	DESCRIPTION
$\bar{E}_a, \bar{E}_b$	Data Output Enable
I <sub>0a</sub> to I <sub>3a</sub>	Data Inputs ( a )
I <sub>0b</sub> to I <sub>3b</sub>	Data Inputs ( b )
O <sub>0a</sub> to O <sub>3a</sub>	Data Outputs ( a )
O <sub>0b</sub> to O <sub>3b</sub>	Data Outputs ( b )
V <sub>CC</sub>	Supply Voltage
GND	System Ground

TRUTH TABLE (A or B buffers)

INPUTS		OUTPUTS
$\bar{E}$	I <sub>0-3</sub>	O <sub>0-3</sub>
L	L	H
L	H	L
H	X	Z

H = logic "1", L = logic "0", X = don't care,  
Z = high impedance



# MD54/74HCT373R

## Octal D-Type Transparent Latch

Preliminary Information

October '83

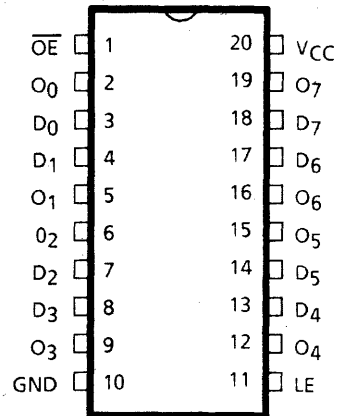
### Features

- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS373 types

### Description

This 8-bit latch features 3-state operation and is designed for use in high speed, bus oriented, systems. The MD54/74HCT373R appears transparent to data (outputs change asynchronously) when Latch Enable (LE) is HIGH. When LE is LOW, data meeting the set up times becomes latched. The state of the latches is unaffected by the active low Output Enable ( $\overline{OE}$ ) pin, but when  $\overline{OE}$  is HIGH the outputs are put into high impedance. Data may thus be latched even when the device is deselected.

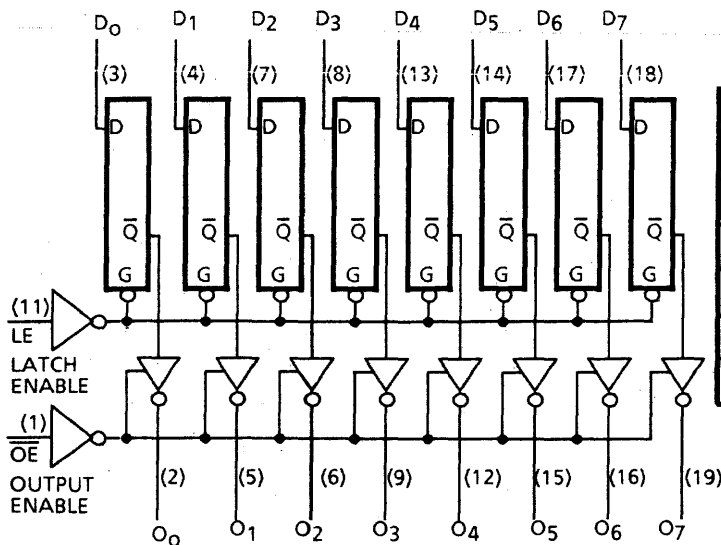
CONNECTION DIAGRAM  
DIP (TOP VIEW)



### Ordering Information

MD54HCT373RCB, Cerdip  
-55°C to 125°C  
with MIL 883B option

MD74HCT373RE, Plastic Dip  
-40°C to 85°C



PIN FUNCTION

PIN	DESCRIPTION
D <sub>0</sub> to D <sub>7</sub>	Data Inputs
O <sub>0</sub> to O <sub>7</sub>	Data Outputs
LE	Latch Enable
$\overline{OE}$	Output Enable
V <sub>CC</sub>	Supply Voltage
GND	System Ground

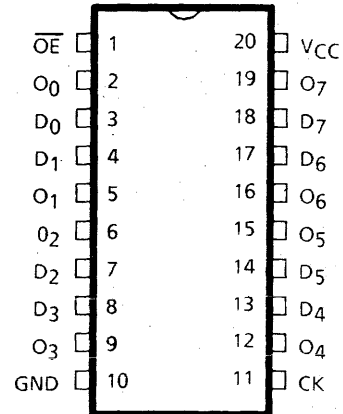
October '83

**Features**

- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS374 types

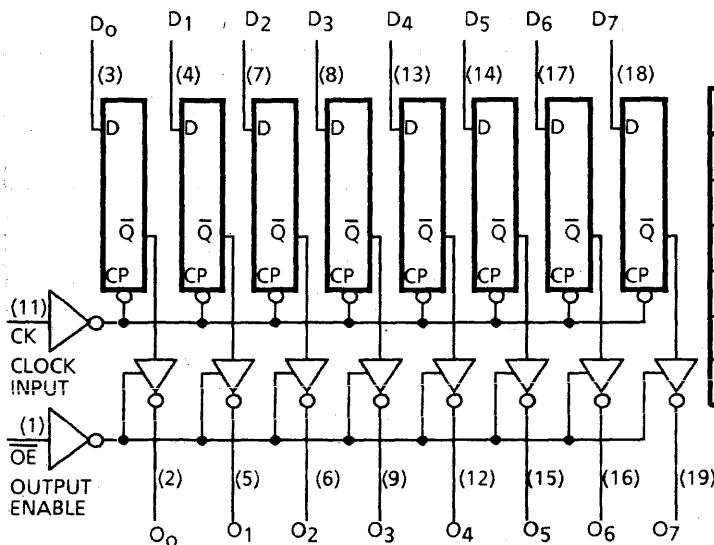
**Description**

This 8-bit latch features 3-state operation and is designed for use in high speed, bus oriented, systems. The latches hold their individual data when meeting set up times with the clock (CK) LOW to HIGH transition. The state of the latches is unaffected by the active low Output Enable ( $\overline{OE}$ ) pin, but when  $\overline{OE}$  is HIGH the outputs are put into high impedance. Data may thus be latched even when the device is deselected.

**CONNECTION DIAGRAM**  
 DIP (TOP VIEW)

**Ordering Information**

MD54HCT374RCB, Cerdip  
 -55°C to 125°C  
 with MIL 883B option

MD74HCT374RE, Plastic Dip  
 -40°C to 85°C


**PIN FUNCTION**

PIN	DESCRIPTION
D <sub>0</sub> to D <sub>7</sub>	Data Inputs
O <sub>0</sub> to O <sub>7</sub>	Data Outputs
CK	Clock Input
$\overline{OE}$	Output Enable
V <sub>CC</sub>	Supply Voltage
GND	System Ground



# MD54/74HCT540R

## Inverting Octal Line Driver/Buffer

Preliminary Information

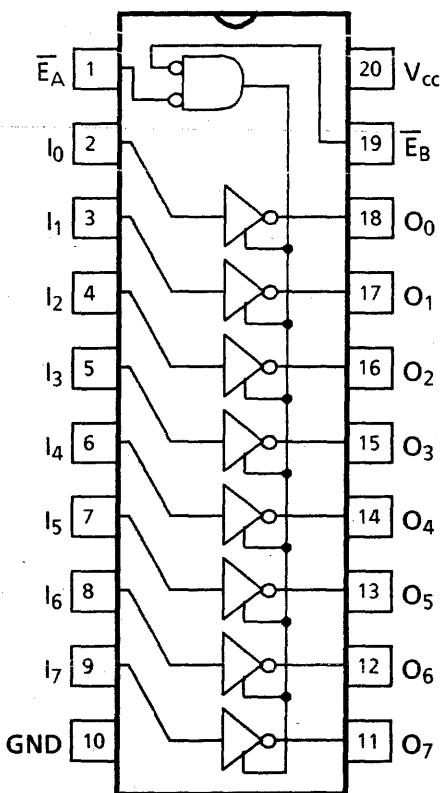
October '83

### Features

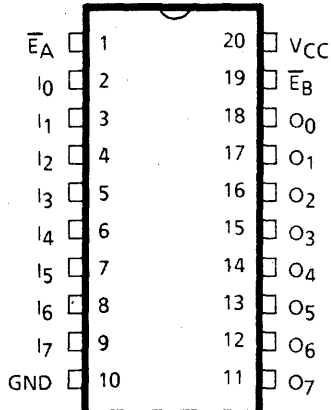
- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS540 types

### Description

These ISO-CMOS Octal line driver/buffers are designed to improve P.C. board density and performance in three-state memory address drivers, clock drivers, and bus oriented receivers/transmitters. The MD54/74HCT540R offers inverting three-state buffers controlled by two active low enable pins.



CONNECTION DIAGRAM  
DIP (TOP VIEW)



### Ordering Information

MD54HCT540RCB, Cerdip  
-55°C to 125°C  
with MIL 883B option

MD74HCT540RE, Plastic Dip  
-40°C to 85°C

PIN	DESCRIPTION
$\bar{E}_A, \bar{E}_B$	Data Output Enable
$I_0$ to $I_7$	Data Inputs
$O_0$ to $O_7$	Data Outputs
$V_{CC}$	Supply Voltage
GND	System Ground

TRUTH TABLE

INPUTS			OUTPUTS
$\bar{E}_A$	$\bar{E}_B$	$I_0$ to $I_7$	$O_0$ to $O_7$
L	L	L	H
L	L	H	L
H	X	X	Z
X	H	X	Z

H = logic "1", L = logic "0", X = don't care,  
Z = high impedance



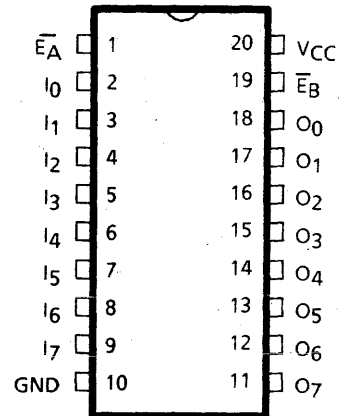
October '83

### Features

- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS541 types

### Description

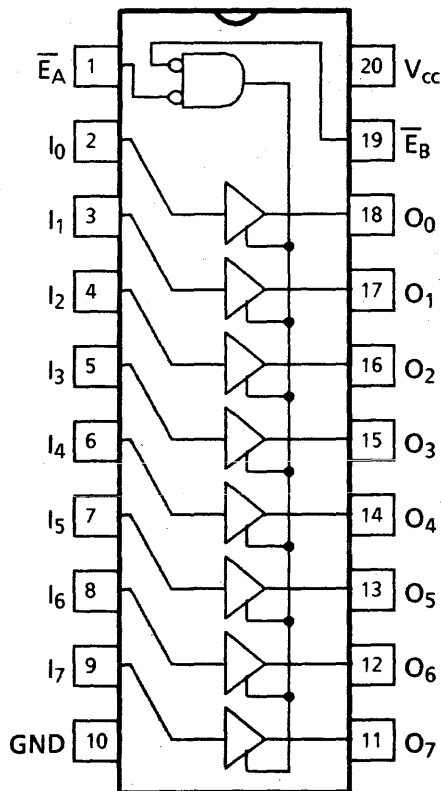
These ISO-CMOS Octal line driver/buffers are designed to improve P.C. board density and performance in three-state memory address drivers, clock drivers, and bus oriented receivers/transmitters. The MD54/74HCT541R offers non-inverting three-state buffers controlled by two active low enable pins.

**CONNECTION DIAGRAM**  
 DIP (TOP VIEW)


### Ordering Information

MD54HCT541RCB, Cerdip  
 -55°C to 125°C  
 with MIL 883B option

MD74HCT541RE, Plastic Dip  
 -40°C to 85°C



PIN	DESCRIPTION
$\bar{E}_A, \bar{E}_B$	Data Output Enable
I <sub>0</sub> to I <sub>7</sub>	Data Inputs
O <sub>0</sub> to O <sub>7</sub>	Data Outputs
V <sub>CC</sub>	Supply Voltage
GND	System Ground

**TRUTH TABLE**

INPUTS			OUTPUTS
$\bar{E}_A$	$\bar{E}_B$	I <sub>0</sub> to I <sub>7</sub>	O <sub>0</sub> to O <sub>7</sub>
L	L	L	L
L	L	H	H
H	X	X	Z
X	H	X	Z

H = logic "1", L = logic "0", X = don't care,  
 Z = high impedance



# MD54/74HCT573R

## Octal D-Type Transparent Latch

Preliminary Information

October '83

### Features

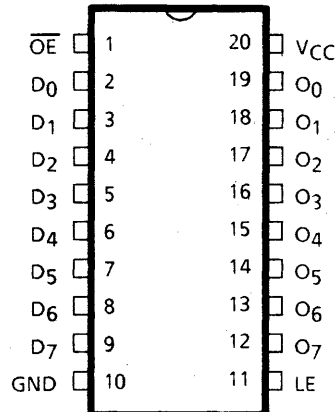
- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS573 types

### Description

This 8-bit latch features 3-state operation and is designed for use in high speed, bus oriented, systems. The MD54/74HCT573R appears transparent to data (outputs change asynchronously) when Latch Enable (LE) is HIGH. When LE is LOW, data meeting the set up times becomes latched. The state of the latches is unaffected by the active low Output Enable ( $\overline{OE}$ ) pin, but when  $\overline{OE}$  is HIGH the outputs are put into high impedance. Data may thus be latched even when the device is deselected.

### CONNECTION DIAGRAM

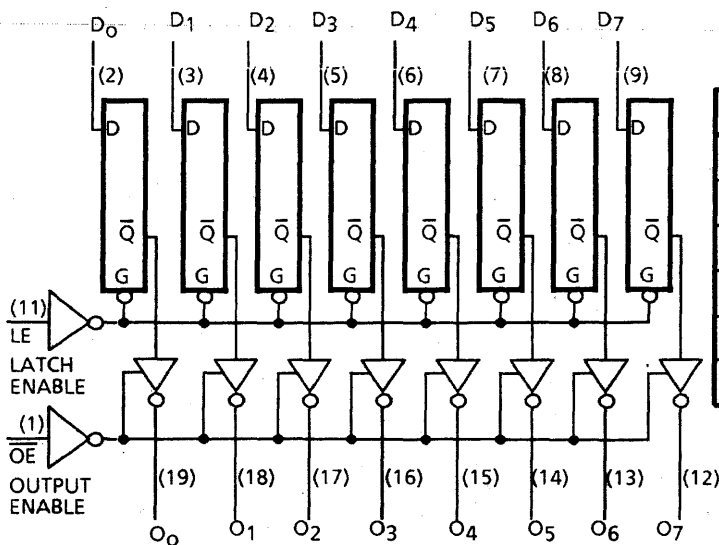
DIP (TOP VIEW)



### Ordering Information

MD54HCT573RCB, Cerdip  
 -55°C to 125°C  
 with MIL 883B option

MD74HCT573RE, Plastic Dip  
 -40°C to 85°C



### PIN FUNCTION

PIN	DESCRIPTION
$D_0$ to $D_7$	Data Inputs
$O_0$ to $O_7$	Data Outputs
LE	Latch Enable
$\overline{OE}$	Output Enable
$V_{CC}$	Supply Voltage
GND	System Ground

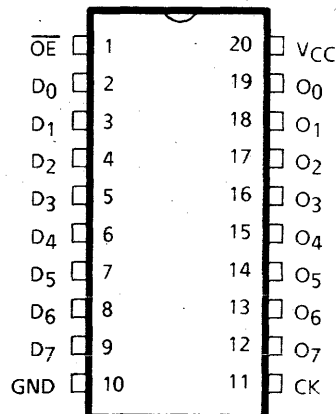
October '83

### Features

- High latch-up immunity
- High current outputs can drive 30 LSTTL loads
- Low power ISO-CMOS technology
- Bus oriented 3-state outputs
- Meets or exceeds all proposed JEDEC 40.2 specifications
- Fully TTL compatible inputs and outputs
- Pin compatible with 54/74LS574 types

### Description

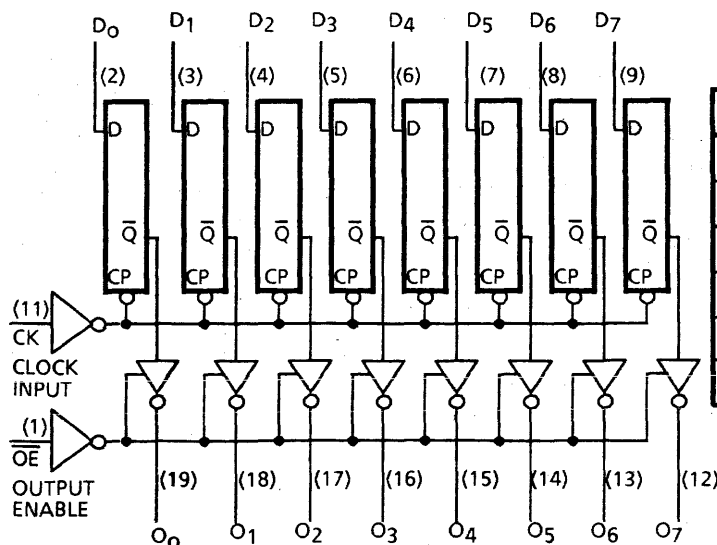
This 8-bit latch features 3-state operation and is designed for use in high speed, bus oriented, systems. The latches hold their individual data when meeting set up times with the clock (CK) LOW to HIGH transition. The state of the latches is unaffected by the active low Output Enable ( $\overline{OE}$ ) pin, but when  $\overline{OE}$  is HIGH the outputs are put into high impedance. Data may thus be latched even when the device is deselected.

**CONNECTION DIAGRAM**  
 DIP (TOP VIEW)


### Ordering Information

MD54HCT574RCB, Cerdip  
 -55°C to 125°C  
 with MIL 883B option

MD74HCT574RE, Plastic Dip  
 -40°C to 85°C


**PIN FUNCTION**

PIN	DESCRIPTION
$D_0$ to $D_7$	Data Inputs
$O_0$ to $O_7$	Data Outputs
CK	Clock Input
$\overline{OE}$	Output Enable
$V_{CC}$	Supply Voltage
GND	System Ground



**MOTOROLA**

# SEMICONDUCTORS

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**MC34010P**  
**MC34011P**

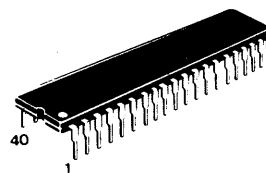
## Advance Information

### ELECTRONIC TELEPHONE CIRCUIT

- Provides All Basic Telephone Station Apparatus Functions in a Single IC, Including DTMF Dialer, Tone Ringer, Speech Network and Line Voltage Regulator
- DTMF Generator Uses Low-Cost Ceramic Resonator with Accurate Frequency Synthesis Technique
- Tone Ringer Drives Piezoelectric Transducer and Satisfies EIA RS-470 Impedance Signature Requirements
- Speech Network Provides Two-Four Wire Conversion with Adjustable Sidetone Utilizing an Electret Transmitter
- On-Chip Regulator Insures Stable Operation Over Wide Range of Loop Lengths
- $\mu 2L$  Technology Provides Low 1.4 Volt Operation and High Static Discharge Immunity
- MC34010P Provides Microprocessor Interface Port for Automatic Dialing Features

**ELECTRONIC  
TELEPHONE  
CIRCUIT**

**BIPOLAR LINEAR/ $\mu 2L$**



**PLASTIC PACKAGE  
CASE 711-03**

## TELECOM SELECTOR GUIDE

### Station Set Products

MC34011	Electronic Telephone Circuit
MC34010	Electronic Telephone Circuit with MPU Interface
MC34012	Tone Ringer
MC34017	Tone Ringer with Push-Pull Output
MC34013	DTMF Generator with Speech Network

### Switching Products

MC3419 Series	Subscriber Loop Interface Circuit
MC3416	4 x 4 x 2 Crosspoint Switch



# MOTOROLA SEMICONDUCTORS

P.O. BOX 20912 • PHOENIX, ARIZONA 85036

## Specifications and Applications Information

### FLOPPY DISK READ AMPLIFIER

The MC3470 is a monolithic READ Amplifier System for obtaining digital information from floppy disk storage. It is designed to accept the differential ac signal produced by the magnetic head and produce a digital output pulse that corresponds to each peak of the input signal. The gain stage amplifies the input waveform and applies it to an external filter network, enabling the active differentiator and time domain filter to produce the desired output.

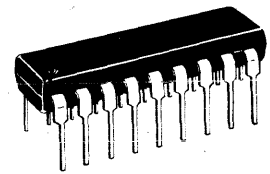
- Combines All the Active Circuitry To Perform the Floppy Disk Read Amplifier Function in One Circuit
- Guaranteed Maximum Peak Shift of 2.0% — MC3470A
- Improved (Positive) Gain  $T_C$  and Tolerance
- Improved Input Common Mode

FLOPPY DISK SUPPORT

## MC3470P MC3470AP

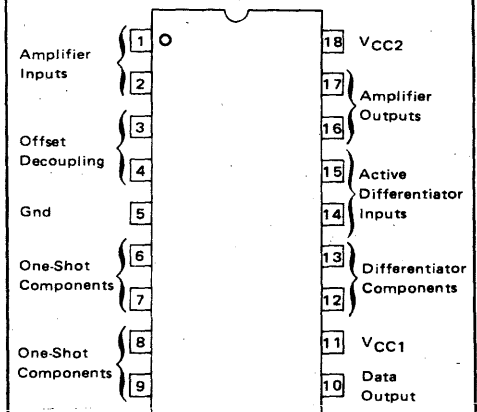
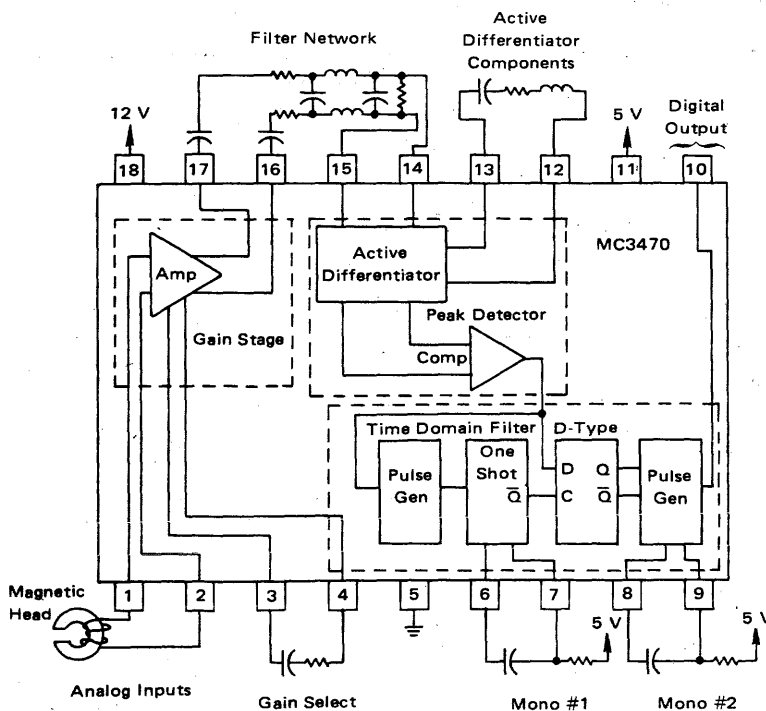
FLOPPY DISK  
READ AMPLIFIER SYSTEM

SILICON MONOLITHIC  
INTEGRATED CIRCUIT



P SUFFIX  
PLASTIC PACKAGE  
CASE 701-01

### TYPICAL APPLICATION



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DS9588  
(Replaces ADI-512)



# MOTOROLA SEMICONDUCTORS

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## Specifications and Applications Information

### FLOPPY DISK WRITE CONTROLLER

The MC3469 and MC3471 are monolithic WRITE Current Controllers designed to provide the entire interface between floppy disk heads and the head control and write data signals for straddle-erase or tunnel-erase heads.

Provisions are made for selecting a range of accurately controlled write currents and for head selection during both read and write operation. Additionally, provisions are included for externally adjusting degauss period and inner/outer track compensation.

The MC3471 provides the erase delay and inhibit functions required to interface with tunnel-erase heads.

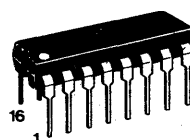
- Head Selection — Current Steering Through Write Head and Erase Coil in Write Mode
- Provides High Impedance (Read Data Enable) during Read Mode
- Head Current (Write) Guaranteed Using Laser Trimmed Internal Resistor (3.0 mA using  $R_{ext} = 10\text{ k}\Omega$ )
- IRW Select Input Provides for Inner/Outer Track Compensation
- Degauss Period Externally Adjustable
- Minimizes External Components
- MC3471 Provides On-Chip Adjustable Erase Delays

### FLOPPY DISK SUPPORT

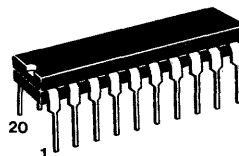
## MC3469P MC3471P

### FLOPPY DISK WRITE CONTROLLERS

#### SILICON MONOLITHIC INTEGRATED CIRCUIT

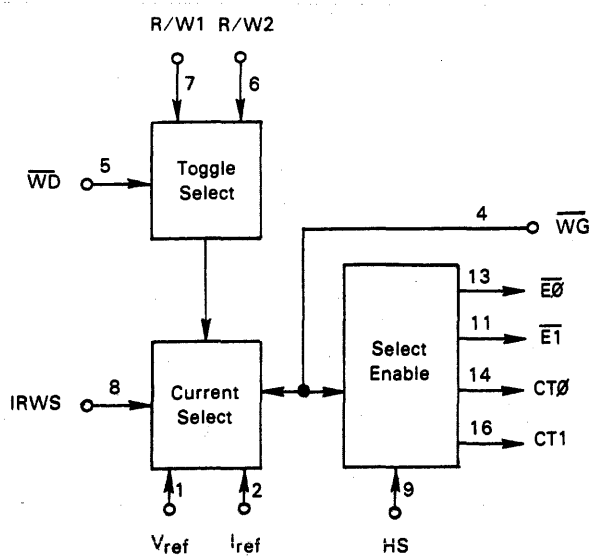


MC3469 only  
P SUFFIX  
PLASTIC PACKAGE  
CASE 648

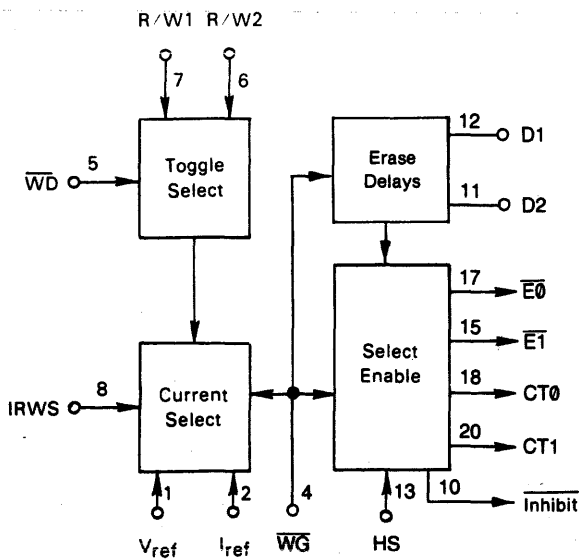


MC3471 only  
P SUFFIX  
PLASTIC PACKAGE  
CASE 738

MC3469  
BLOCK DIAGRAM



MC3471  
BLOCK DIAGRAM





MOTOROLA

# SEMICONDUCTORS

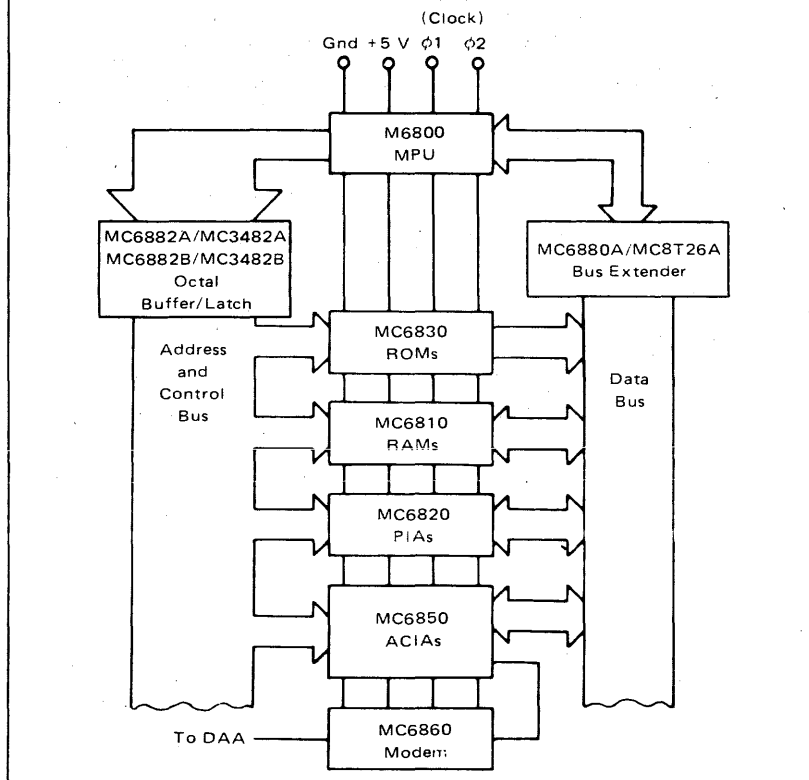
P.O. BOX 20912 • PHOENIX, ARIZONA 85036

## OCTAL THREE-STATE BUFFER/LATCH

This series of devices combines four features usually found desirable in bus-oriented systems: 1) High impedance logic inputs insure that these devices do not seriously load the bus; 2) Three-state logic configuration allows buffers not being utilized to be effectively removed from the bus; 3) Schottky technology allows for high-speed operation; 4) 48 mA drive capability.

- Inverting and Non-Inverting Options of Data
- SN74S373 Function Pinouts
- Eight Transparent Latches/Buffers in a Single Package
- Full Parallel-Access for Loading and Reloading
- Buffered Control Inputs
- All Inputs Have Hysteresis to Improve Noise Rejection
- High Speed – 8.0 ns (Typ)
- Three-State Logic Configuration
- Single +5 V Power Supply Requirement
- Compatible with 74S Logic or M6800 Microprocessor Systems
- High Impedance PNP Inputs Assure Minimal Loading of the Bus

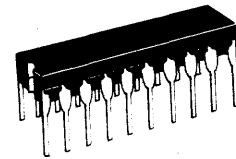
## MICROPROCESSOR BUS EXTENDER APPLICATION



## MC3482A/MC6882A MC3482B/MC6882B

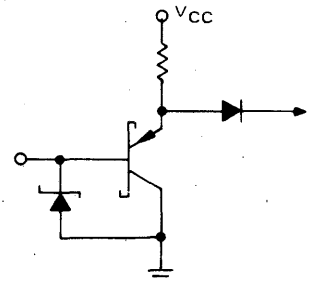
This device may be ordered under either of the above type numbers.

## OCTAL THREE-STATE BUFFER/LATCH

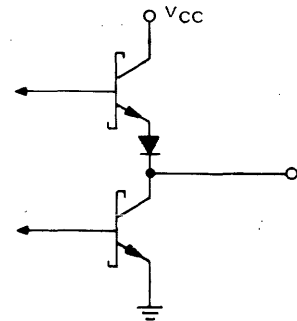


L SUFFIX  
CASE 732

## INPUT EQUIVALENT CIRCUIT



## OUTPUT EQUIVALENT CIRCUIT



## ORDERING INFORMATION

(Temperature Range for the following devices = 0 to +75°C.)

Device	Alternate	Package
MC6882AL	MC3482AL	Ceramic DIP
MC6882BL	MC3482BL	Ceramic DIP

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MOTOROLA BROADLINE

## BUS STANDARD LINE DRIVERS & RECEIVERS

STANDARD	INTRODUCED	1983 INTRODUCTION
RS232C	MC1488 DR MC1489 REC	
RS422	MC3486 REC MC3487 DR AM26LS31 DR	AM26LS32REC
RS423	MC3487 DR	$\mu$ A9636A (MC3488A)
RS-485 (PARTY LINE)	SN75172 REC SN75174 REC SN75173 DR SN75175 DR	SN75176-8
IEEE 488-1978	MC3440-43A MC3446A IMC3448A MC3447	
IBM GA-22-6974-3	MC3481 MC3485 MC75125-9	

## PERIPHERAL DRIVERS

PART NUMBER	DESCRIPTION	INPUT COMPATIBILITY				
MC1411P (ULN2001A) MC1412P (ULN2002A) MC1413P (ULN2003A) MC1416P (ULN2004A) MC1472U/P1	<table style="border: none;"> <tr> <td style="font-size: 2em; vertical-align: middle;">}</td> <td>Seven Element Darlington Array (0.5 A Max)</td> </tr> <tr> <td style="font-size: 2em; vertical-align: middle;">}</td> <td>Dual Positive NAND</td> </tr> </table>	}	Seven Element Darlington Array (0.5 A Max)	}	Dual Positive NAND	General PMOS TTL/CMOS NMOS H1 Z(1)
}	Seven Element Darlington Array (0.5 A Max)					
}	Dual Positive NAND					
ULN2801A ULN2802A ULN2803A ULN2804A ULN2068B ULN2074B	<table style="border: none;"> <tr> <td style="font-size: 2em; vertical-align: middle;">}</td> <td>Octal Darlington Array (0.5 A Max)</td> </tr> <tr> <td></td> <td>Quad 1.5 A Darlington<sup>(2)</sup> Quad 1.5 A Darlington<sup>(3)</sup></td> </tr> </table>	}	Octal Darlington Array (0.5 A Max)		Quad 1.5 A Darlington <sup>(2)</sup> Quad 1.5 A Darlington <sup>(3)</sup>	General PMOS TTL/CMOS CMOS/PMOS TTL/CMOS/DTL General
}	Octal Darlington Array (0.5 A Max)					
	Quad 1.5 A Darlington <sup>(2)</sup> Quad 1.5 A Darlington <sup>(3)</sup>					

- (1) PNP Bufferer for TTL/MOS Compatibility
- (2) Open collector — Diode Clamped
- (3) Open Collector — Open Emitter





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# SEMICONDUCTORS

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## MICROPROCESSOR/COMPUTER INTERFACE MC3481 MC3485

### QUAD SINGLE-ENDED LINE DRIVER

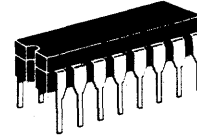
The MC3481 and MC3485 are quad single-ended line drivers specifically designed to meet the IBM 360/370 I/O specification (GA22-6974-3).

Output levels are guaranteed over the full range of output load and fault conditions. Compliance with the IBM requirements for fault protection, flagging, and power up/power down protection for the bus make this an ideal line driver for party line operations.

- Separate Enable and Fault Flags — MC3481
- Common Enable and Fault Flag — MC3485
- Power Up/Down Does Not Disturb Bus
- Schottky Circuitry for High-Speed — PNP Inputs
- Internal Bootstraps for Faster Rise Times
- Driver Output Current Foldback Protection
- MC3485 has LS Totem Pole Driver Output

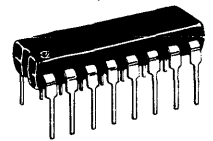
### IBM 360/370 QUAD LINE DRIVER

SILICON MONOLITHIC  
INTEGRATED CIRCUIT

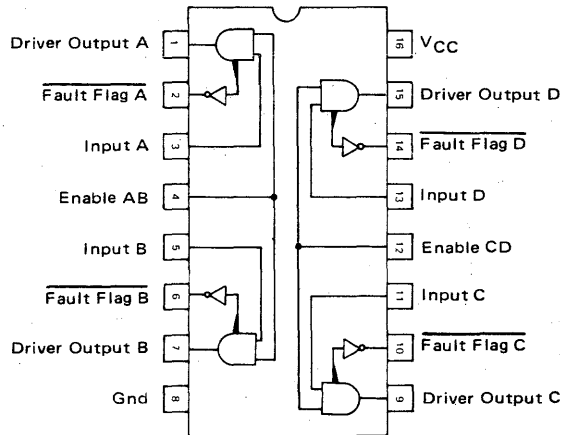


**L SUFFIX**  
CERAMIC PACKAGE  
CASE 620

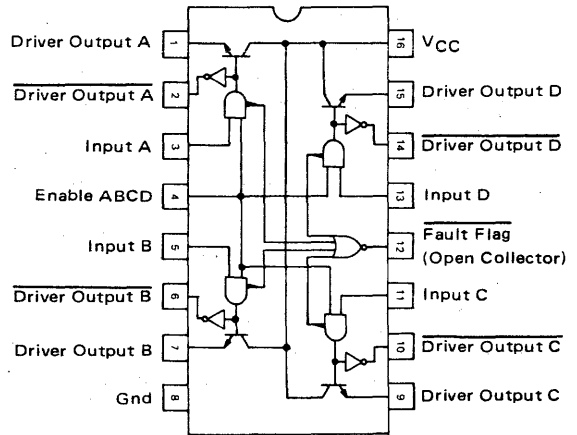
**P SUFFIX**  
PLASTIC PACKAGE  
CASE 648



**MC3481  
DUAL ENABLE  
INDIVIDUAL FAULT FLAG**



**MC3485  
COMMON ENABLE  
COMMON FAULT FLAG**



### SUPPORTING IBM 360/370 LINE RECEIVERS

Device Number	No. of Channels	Input Resistance kΩ Min/Max	$I_{IH(R)}$ @ $V_{IH} = 3.11$ V mA Max	$t_{PLH}$ @ $C_L = 50$ pF ns Max
MC75125/75127	Seven	7.4/20	0.42	25
MC75128/75129	Eight	7.4/20	0.42	25

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# MOTOROLA SEMICONDUCTORS

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## BIDIRECTIONAL INSTRUMENTATION BUS (GPIB) TRANSCEIVER

This bidirectional bus transceiver is intended as the interface between TTL or MOS logic and the IEEE Standard Instrumentation Bus (488-1978, often referred to as GPIB). The required bus termination is internally provided.

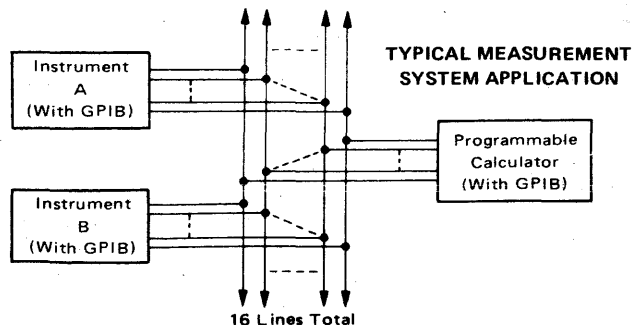
Low power consumption has been achieved by trading a minimum of speed for low current drain on non-critical channels. A fast channel is provided for critical ATN and EOI paths.

Each driver/receiver pair forms the complete interface between the bus and an instrument. Either the driver or the receiver of each channel is enabled by a Send/Receive input with the disabled output of the pair forced to a high impedance state. The receivers have input hysteresis to improve noise margin, and their input loading follows the bus standard specifications.

- Low Power — Average Power Supply Current = 30 mA Listening  
75 mA Talking
- Eight Driver/Receiver Pairs
- Three-State Outputs
- High Impedance Inputs
- Receiver Hysteresis — 600 mV (Typ)
- Fast Propagation Times — 15–20 ns (Typ)
- TTL Compatible Receiver Outputs
- Single +5 Volt Supply
- Open Collector Driver Output with Terminations
- Power Up/Power Down Protection (No Invalid Information Transmitted to Bus)
- No Bus Loading When Power is Removed From Device
- Required Termination Characteristics Provided

### MAXIMUM RATINGS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

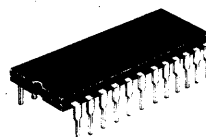
Rating	Symbol	Value	Unit
Power Supply Voltage	$V_{CC}$	7.0	Vdc
Input Voltage	$V_I$	5.5	Vdc
Driver Output Current	$I_{O(D)}$	150	mA
Junction Temperature	$T_J$	150	$^\circ\text{C}$
Operating Ambient Temperature Range	$T_A$	0 to +70	$^\circ\text{C}$
Storage Temperature Range	$T_{stg}$	-65 to +150	$^\circ\text{C}$



## GPIB INSTRUMENTATION BUS INTERFACE MC3447

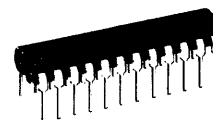
### OCTAL BIDIRECTIONAL BUS TRANSCEIVER WITH TERMINATION NETWORKS

SILICON MONOLITHIC INTEGRATED CIRCUIT

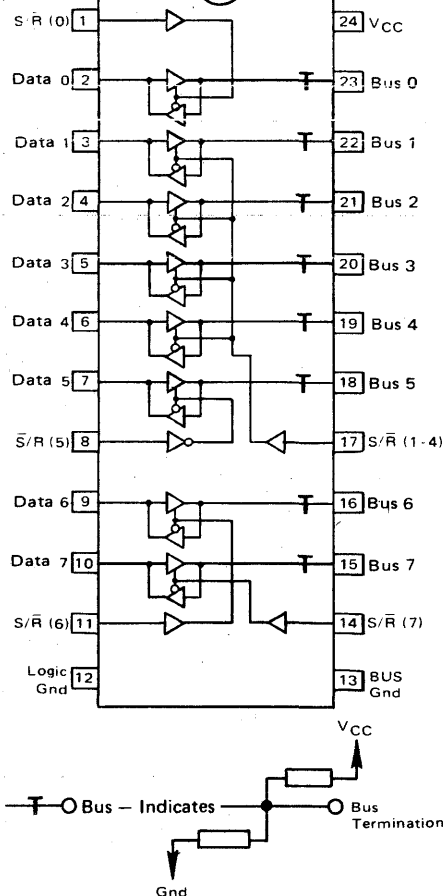


**L SUFFIX**  
CERAMIC PACKAGE  
CASE 623

**P3 SUFFIX**  
PLASTIC PACKAGE  
CASE 724



### PIN ASSIGNMENTS



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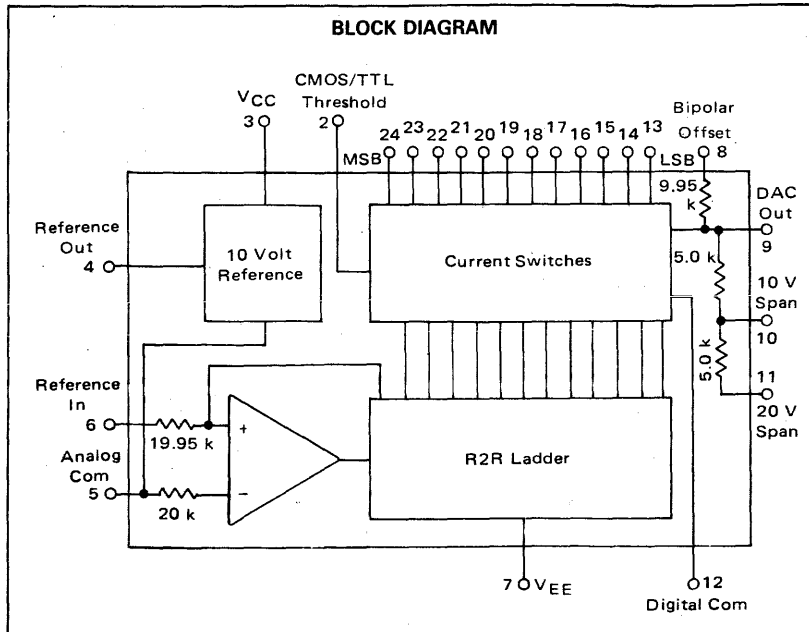
## Advance Information

### COMPLETE 12-BIT HIGH-SPEED MONOLITHIC D/A CONVERTER

The MC3412/3512 is a monolithic 12-bit resolution D/A converter. It contains a high stability bandgap reference capable of supplying 1.5 mA externally, trimmed to  $\pm 0.25\%$  maximum error. Active laser trimming of thin film ladder network, reference, span, bipolar offset, and bandgap resistors at wafer level provide accuracy and linearity of better than  $\pm 1/2$  LSB. An innovative bit switching scheme provide fast settling time yet enables selection of CMOS or TTL thresholds. TTL threshold levels are retained over a wide VCC range from 13.5 to 16.5 volts. Precision internal span resistors allow output voltage options of 0 to 5.0 V, 0 to 10 V,  $\pm 2.5$  V,  $\pm 5.0$  V, and  $\pm 10$  V. 12-bit accuracy and a fast settling time of typically 200 ns (to  $\pm 1/2$  LSB) make this converter ideal for applications such as a fast A/D building block or display driver.

- True 12-Bit Linearity:  $\pm 1/2$  LSB Max
- Fast Settling Time:  $\pm 1/2$  LSB in 200 ns Typ.
- Fully Monotonic Over Temperature Range
- High-Stability Bandgap Voltage Reference On Chip
- Linearity Guaranteed Over Temperature
- Low Power Consumption: 210 mW
- Pinout Compatible with AD563 and AD565
- Selectable Digital Thresholds
- Internal Span Resistors for Generating Output Voltage

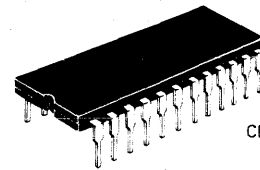
### BLOCK DIAGRAM



## DATA CONVERSION MC3412 MC3512

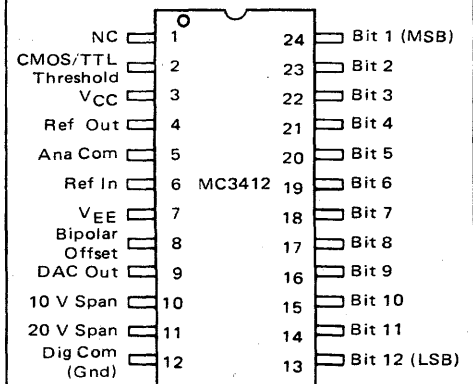
### LASER TRIMMED HIGH-SPEED 12-BIT D/A CONVERTER

SILICON MONOLITHIC  
INTEGRATED CIRCUIT



**L SUFFIX**  
CERAMIC PACKAGE  
CASE 623

### PIN CONNECTIONS



### ORDERING INFORMATION

Device	Temperature Range	Package
MC3412L	0°C to +70°C	Ceramic DIP
MC3512L	-55°C to +125°C	Ceramic DIP

This is advance information and specifications are subject to change without notice.

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**MOTOROLA**

# SEMICONDUCTORS

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## Advance Information

### MPU-BUS-COMPATIBLE 8-BIT D-TO-A CONVERTER

The MC6890 is a self-contained, bus-compatible, 8 bit ( $\pm 0.19\%$  accuracy) D-to-A converter system capable of interfacing directly with 8-bit microprocessors.

Available in both commercial and military temperature ranges, this monolithic converter contains master/slave registers to prevent transparency to data transitions during active enable; a laser-trimmed, low-TC, 2.5 V precision bandgap reference; and high stability, laser-trimmed, thin-film resistors for both reference input and output span and bipolar offset control.

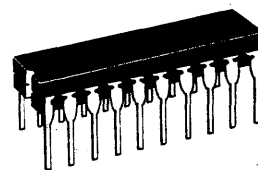
A reset pin provides for overriding stored data and forcing  $I_{out}$  to zero.

- Direct Data Bus Link with All Popular TTL Level MPU's
- $\pm 1/2$  LSB Nonlinearity Over Temperature
- Fast Settling Time: 200 ns Typ
- Internal 2.5-V Precision Laser-Trimmed Voltage Reference (May Also Be Used Externally)
- Minimum  $\overline{\text{Enable}}$  Pulse Width: 70 ns
- Fast  $\overline{\text{Enable}}$ : 10 ns Maximum Data Hold Time
- $\overline{\text{Reset}}$  Pin to Override Data
- Output Voltage Ranges: +5, +10, +20, or  $\pm 2.5$ ,  $\pm 5$ ,  $\pm 10$  Volts
- Low Power: 90 mW Typ
- +5 V and -5 V to -15 V Supplies

## DATA CONVERSION MC6890

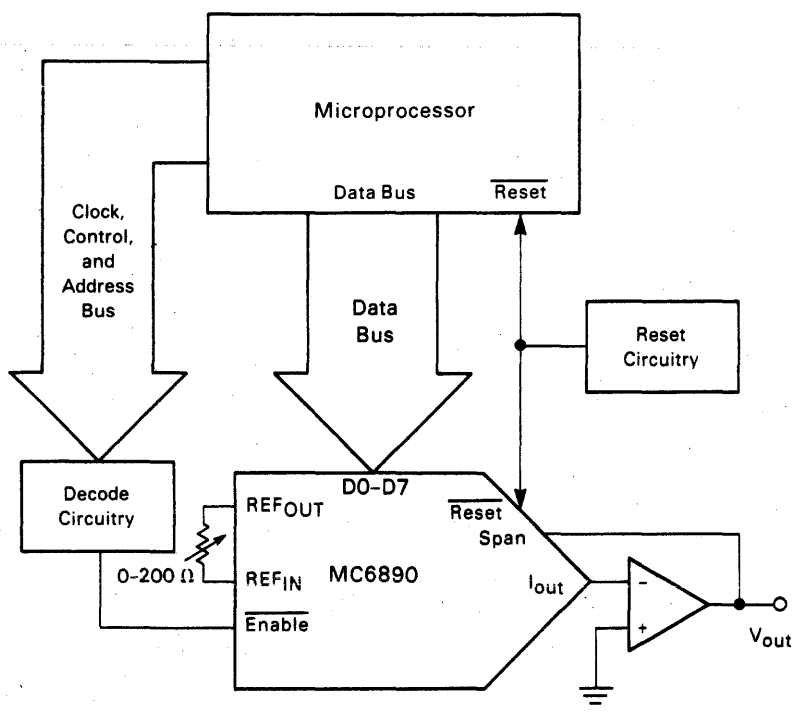
### 8-BIT MPU-BUS-COMPATIBLE DAC

SILICON MONOLITHIC  
INTEGRATED CIRCUIT

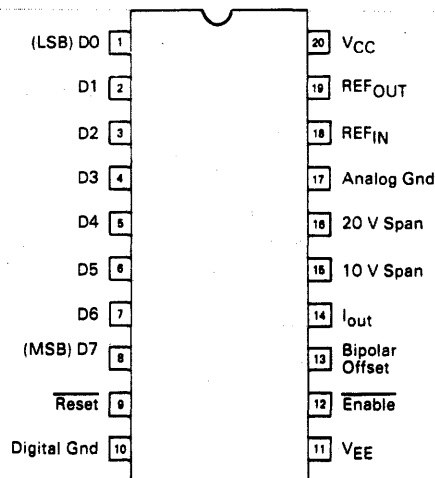


L SUFFIX  
CASE 732

### OPERATION WITH AN MPU



### PIN CONNECTIONS



### ORDERING INFORMATION

Device	Temperature Range	Package
MC6890L	0° to +70°C	Ceramic DIP
MC6890AL	-55° to +125°C	Ceramic DIP

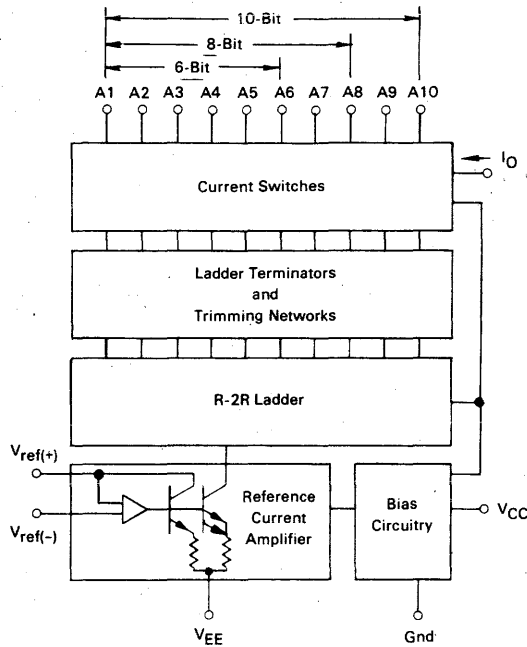
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ADI-592R2

This document contains information on a new product. Specifications and information herein are subject to change without notice.

# ADDITIONAL D-A CONVERTERS

## General Purpose



Device Number	Error (%Max)	$P_D$ @ $V_{EE} = -5V$ (mW Max)	$t_{Settling}$ (ns Typ)	$I_O$ (mA)	Suffix	Case
<b>6-Bit</b>						
MC1506* MC1406	$\pm 0.78$	120	150	1.9 to 2.1	L	632
<b>8-Bit</b>						
DAC-08A DAC-08* DAC-08H DAC-08E DAC-08C	$\pm 0.1$ $\pm 0.19$ $\pm 0.1$ $\pm 0.19$ $\pm 0.39$	48	85	2.984 to 2.880 1.940 to 2.040	Q P, Q	620 648, 620
MC1508L8* MC1408L8	$\pm 0.19$				L	620
MC1408L7 MC1408L6 MC3408	$\pm 0.39$ $\pm 0.78$ $\pm 0.5$	170	300	1.9 to 2.1	L, P L	620, 648 620
<b>10-Bit</b>						
MC3510* MC3410 MC3410C	$\pm 0.05$ $\pm 0.1$	220	250	3.8 to 4.2	L L, P	690 690, 648
<b>12-Bit</b>						
AD562SD* AD562AD # AD562KD	$\pm 0.006$ $\pm 0.01$	210	200	1.6 to 2.4	D	716
AD563TD* AD563SD* AD563KD AD563JD	$\pm 0.006$ $\pm 0.01$	210	200	1.6 to 2.4	D	716

\* $T_A = -55$  to  $+125^\circ C$  # $T_A = -25$  to  $+85^\circ C$  Devices without symbol:  $T_A = 0$  to  $+70^\circ C$

**Bold Part Numbers Indicate New Products.**

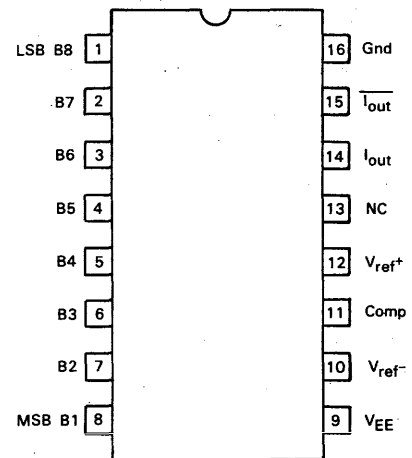
## D-A CONVERTERS — HIGH SPEED

MC10318 series — A high speed 8-bit D/A converter capable of data conversion rates in excess of 25.MHz. It is intended for applications in high speed instrumentation and communication equipment, graphics displays, storage oscilloscopes, radar processing, and TV broadcast systems. The inputs are compatible with MECL 10,000 series logic, while the complementary current outputs have 51 mA full scale capability. Monotonic over the full temperature range, the outputs typically settle in 10 ns.

$T_A = 0$  to  $70^\circ C$

Packages:

L Suffix — Case 620/690



Device Number	Error % Max	$P_D$ @ $V_{EE} = -5.2V$ mW Max	$t_{Settling}$ ns Typ (to $\frac{1}{2}$ LSB)	$I_O$ & $\bar{I}_O$ @ $V_{Ref} = 10.56V$ mA Typ
<b>MC10318CL6</b> <b>MC10318CL7</b>	$\pm 0.78$ $\pm 0.39$	675 675	10 10	51 51
MC10318L MC10318L9	$\pm 0.19$ $\pm 0.10$	675 675	10 10	51 51

**Bold Part Numbers Indicate New Products.**



# MOTOROLA SEMICONDUCTORS

P.O. BOX 20912 • PHOENIX, ARIZONA 85036

## Advance Information

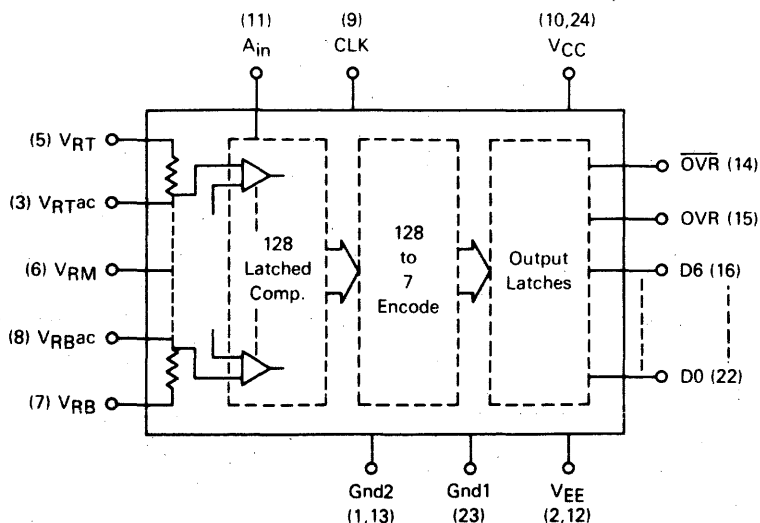
### SEVEN-BIT PARALLEL HIGH SPEED A/D CONVERTER (WITH OVERRANGE)

The MC10315L/MC10317L is a 7-bit high speed parallel A/D converter which employs ECL processing. The device consists of 128 parallel latched comparators across a high quality input reference network. The 128 comparator outputs are then fed to a 128-to-7 encoder and latched to the outputs which are ECL compatible. An overrange bit is provided to allow overrange sensing, or to facilitate the connection of an MC10315L and MC10317L in parallel to produce an 8-bit A/D converter. The MC10315L and MC10317L are identical devices except for the method of overranging used, which simplifies the utilization of two 7-bit converters to produce an 8-bit conversion. (See ordering information and technical description.)

Applications include video display and radar signal processing, high speed instrumentation, and TV broadcast video encoding.

- 7-Bit Resolution/8-Bit Accuracy Plus Overrange
- Direct Interconnection for 8-Bit Conversion
- 15 MHz Sampling Rate
- Wide Range of Input Voltage:  $\pm 2.0$  Volts
- Low Input Capacitance:  $\leq 70$  pF
- 1.2 Watt Power Dissipation
- No Sample and Hold Required for Video Bandwidth Signals
- Standard 24-Pin Package

### MC10315L/MC10317L DEVICE/APPLICATION CONFIGURATION



This document contains information on a new product. Specifications and information herein are subject to change without notice.

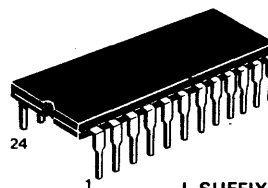
© MOTOROLA INC., 1982

## DATA CONVERSION

# MC10315L MC10317L

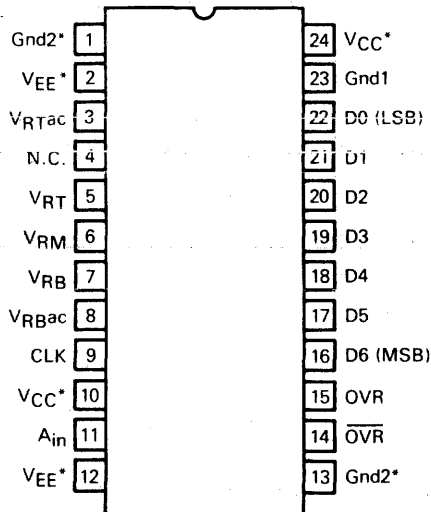
### HIGH SPEED 7-BIT ANALOG-TO-DIGITAL FLASH CONVERTER

SILICON MONOLITHIC  
INTEGRATED CIRCUIT



L SUFFIX  
CERAMIC PACKAGE  
CASE 623

### PIN DIAGRAM



\*VCC, VEE and Gnd2 are each available on two pins. Interconnections for the respective function are made on chip. To minimize I<sup>2</sup>R drops on chip and in the bonding wires, utilization of both pins for each function is recommended.

### ORDERING INFORMATION\*\*

Device	Overrange Function		
	Analog Input Condition	Logic Levels	
		OVR Bit	D0-D6 Bits
MC10315L	Overranged	High	High
MC10317L	Overranged	High	Low

\*\*For information regarding an evaluation board, contact Linear Marketing.

ADI-654R1

# SP9770

## TEN BIT HIGH SPEED D TO A CONVERTER

The Plessey SP9770 10-bit D/A converter is capable of converting a digital signal into an analog voltage at a rate of over 75 mega-bits per second (MBPS). An inherently low glitch desing is used and the complementary current outputs are suitable for direct transmission line drive. Included on the chip is a high performance voltage reference and reference amplifier.

### FEATURES

- 12ns Settling time
- 10 bits to  $\pm 1/2$  LSB absolute and differential non linearity
- 75 MBPS Update rate
- Current output
- ECL standard inputs
- Complementary outputs, 20 mA full scale
- Reference tempco. typically 40 ppm/ $^{\circ}$ C
- DAC usable in multiplying mode to 25MHz
- 24 lead, 0.6 pitch package

### ABSOLUTE MAXIMUM RATINGS

Positive supply voltage	+ 5.5V
Negative supply voltage	-5.7V
Operating temperature range	-30 $^{\circ}$ C to 85 $^{\circ}$ C
Storage temperature	-55 $^{\circ}$ C to 125 $^{\circ}$ C
Lead temperature (soldering 60 sec)	300 $^{\circ}$ C
Digital input voltage	0 to VEE
Min current set resistance	R SET 150 $\Omega$
Output reference supply	0 to +1V

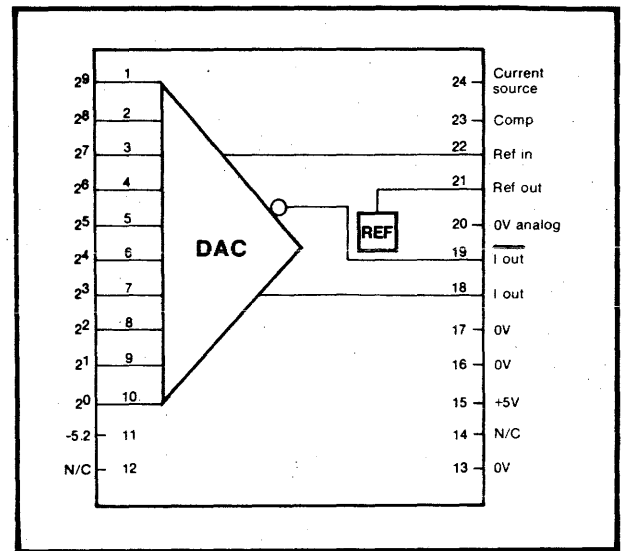


Fig. 1 SP9770 pin connections

### OPERATING NOTES

The pinout of the device is shown in Fig. 1. External components are limited to the current setting resistor and decoupling capacitors. The DAC has current outputs, with a nominal full scale of 20mA, corresponding with a 1 volt drop across a 50 ohm load. The actual output current is determined by the on-chip reference voltage and an off-chip current setting resistor. Output current, I out, is given by

$$I_{OUT} = 4 \times \frac{V_{REF}}{R_{SET}}$$

A complementary Iout is also provided. The setting resistor, RSET, is typically 240 ohms, giving a full scale output current of 20.75mA, and should have a temperature coefficient similar to that of the output load resistor.

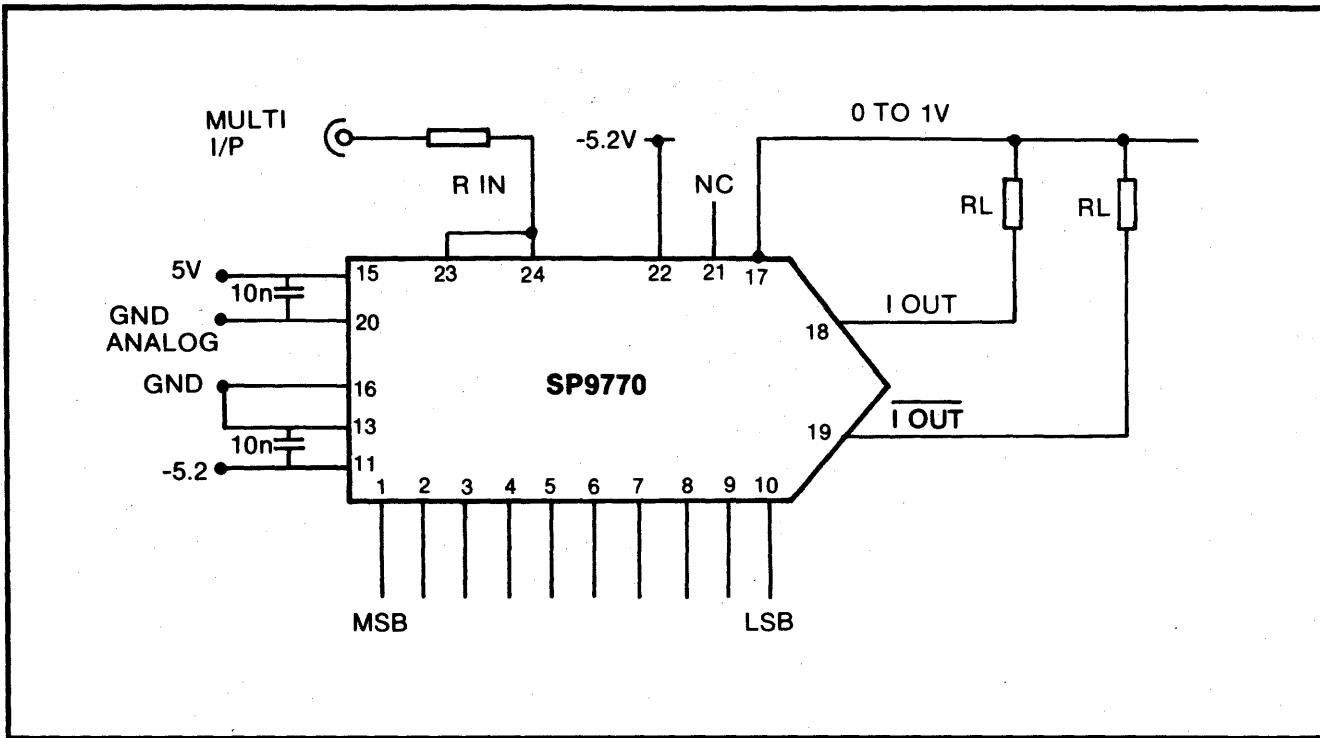


Figure 4. Multiplying Mode operation

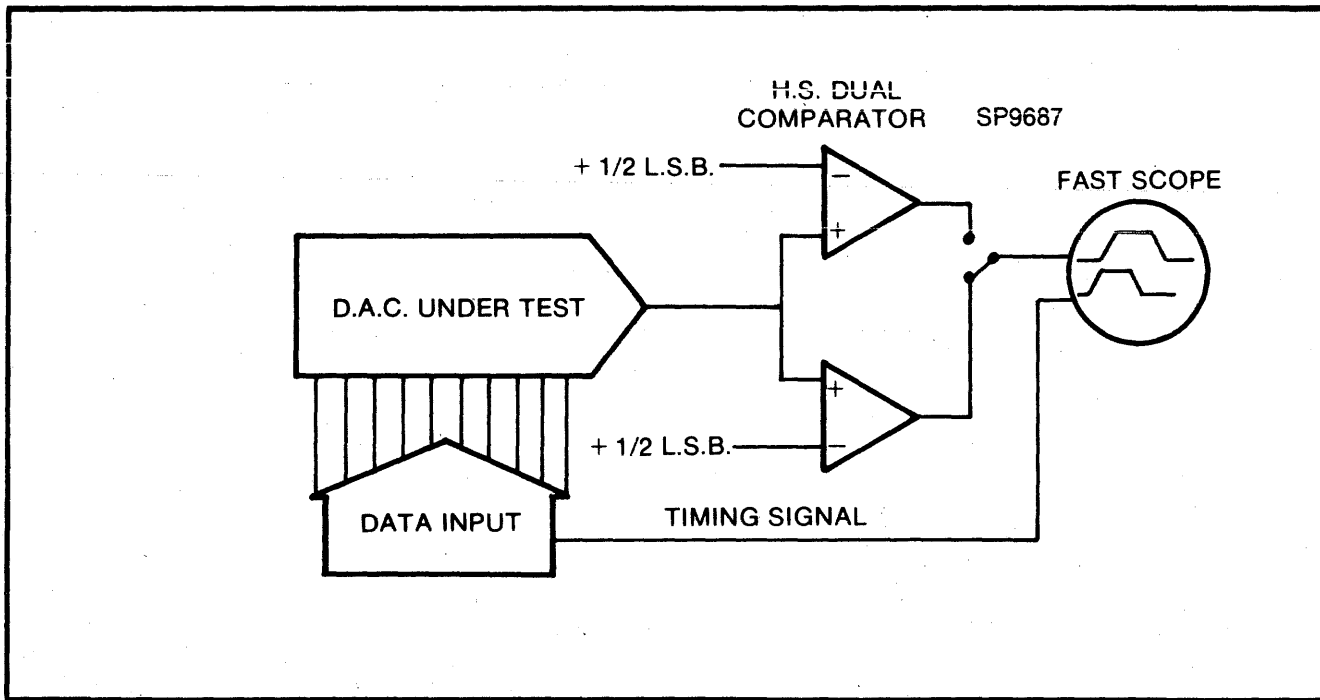


Figure 5. Window detection of D.A.C. output settling



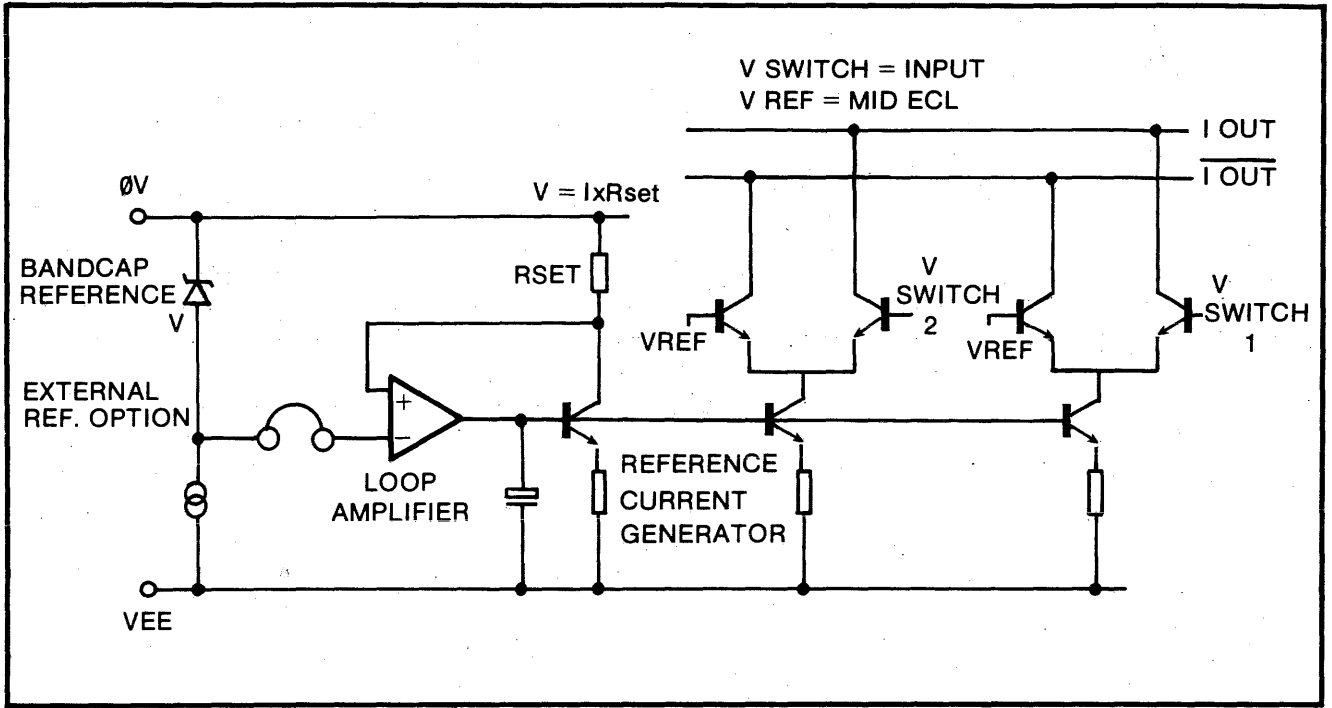


Figure 2. Basic Current Switched DAC

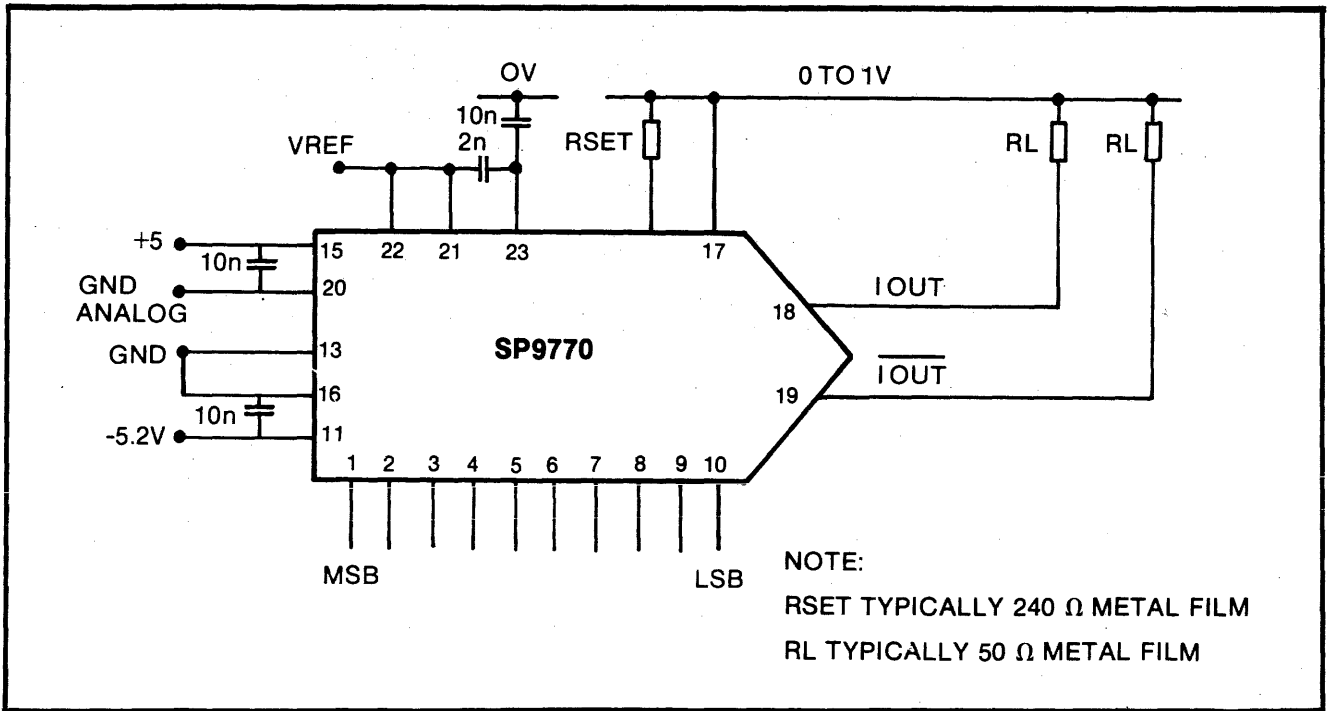


Figure 3. Conventional D/A operation using on-chip reference

The SP9770 has been designed to meet the dc specification shown in the test table below.

Parameter	Symbol	Test Limits			Unit	Comments
		Min.	Typ.	Max.		
Power Supply Current	ICC IEE	7 45	12 56	17 70	mA dc	
Digital Input Currents	IIN H	—	115	200	μAmps	All Inputs High
Reference Voltage	VREF	-1.250	-1.280	-1.300	Volts	
Reference Voltage Temp Coefficient	ΔVREF/ΔT	—	40	80	PPM/°C	
Resolution		—	0.098 10	— —	% Bits	
Non-Linearity	Differential Absolute	— —	— —	±0.05 ±0.05	% %	
Bit Size		19 0.9	21 1	23 1.1	μAmps mVolt	RL = 50Ω
Fullscale Settling Time	ts	—	12	20	nSec	To 1 LSB
Multiplying Bandwidth (3dB)	BM	—	25	—	MHz	Current Mode

### RECOMMENDED OPERATING CONDITIONS

Parameter	Symbol	Test Limits			Unit	Comment
		Min.	Typ.	Max.		
Power Supply Voltage	VEE VCC	-4.85 +4.75	-5.3 +5.0	-5.45 +5.25	VOLTS dc	
Input Logic Levels	VIH VIL	-0.960 -1.850	-- --	-0.810 -1.650	VOLTS VOLTS	ECL. 10k Compatible
Fullscale o/p Current	Iout	2	20	30	mAmps	$I_{out} = \frac{4 \text{ REF}}{R_{Set}}$
Output Compliance Voltage (at Pin Under Test) (See Note 1.)	Vc -1	-0.7 --	-- +1	+1	VOLTS VOLTS	at 85°C at 25°C
Ambient Operating Temperature	T	-30	--	+85	°C	

#### Note 1

Pin 17 should be connected to a reference supply at the most positive voltages required, with the load and Rset pin 24 returned similarly to this reference. Conveniently the reference is ground, but this limits the output to negative excursions only. Refer to figure 3.



The reference voltage source is nominally — 1.28 volts and is of a modified bandgap type. Samples show average temperature coefficients of 40ppm/°C over the range — 55°C to 125°C.

The reference supply is internally compensated; however, to reduce the possibility of instability or noise generation, pin 21 should be decoupled to pin 23 with 2nF ceramic chip capacitor. The current loop technique (Fig 2) has been used with a high performance loop amplifier. The current is set by an external resistor as described above. Stabilization of the loop amplifier is achieved by a single capacitor from pin 23 to ground. Minimum value is 3900pF, although a 20nF chip ceramic is recommended.

Multiplying operation of the DAC is available in two modes, either a voltage applied in place of the reference, or a current supplied via the current source pin. In the former case the 3dB bandwidth is 250kHz, while in the latter, operational use is typically 25MHz. A suggested circuit is shown in Fig. 4.

## ELECTRICAL CHARACTERISTICS

Test conditions (unless otherwise stated)

$$T = 25^{\circ}\text{C}, R_{\text{set}} = 240\Omega$$

VIH MAX	VIL MIN	VEE	VCC
-0.81	-1.85	-5.2	+5.0

## Measurement of Settling Time

Oscilloscopes, whether real time or sampling do not have sufficiently low input VSWR or on-screen resolution for precise settling time measurements. A measurement technique has been devised, shown diagrammatically in Fig. 4, in which the DAC can settle into nearly ideal 50 ohm load, with minimal interconnection paths; this also very closely related to the practical use of the device. Precision settling time measurements can be performed with a high speed comparator, conveniently a dual device, such as the SP9687, with a minimal delay time, in this case about 2ns. Two references are set up to detect the DAC output settling within a window, conveniently defined as the settling to ground of the output.

The lower comparator detects when the DAC output has settled within 1/2 LSB of the final settling point, while the upper device when switched in checks that there is no overshoot. In fact, the settling is very well behaved, and after correction for comparator delay, the results consistently show a DAC settling time of 12ns, defined from the 50% point of the DAC input switching waveform. No correction has been made for the excess loading capacitance at the output node, so these results are considered to be conservative.



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# Analog Switch Product Information

## Drivers and Gates

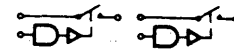
Basic Part No. (Notes 1 & 2)	Switch Type	r <sub>DS(on)</sub> Max. (Ω) (Note 3)	Analog Voltage Range (p-pV) (Note 3)	Switching Time (μs)		Logic Input for ON Switch	Logic Levels (V)		Opt. Sup. Voltage (V)			Comments
				t <sub>ON</sub>	t <sub>OFF</sub>		V <sub>INL</sub>	V <sub>INH</sub>	(+) Sup. V <sub>1</sub>	(-) Sup. V <sub>2</sub>	Logic Sup. V <sub>I</sub>	



One Channel SPST

### One Channel SPST

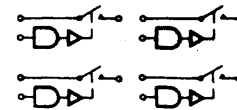
DG5040	CMOS Plus-40	50	30	1.0	0.5	1	0.8	2.0	15	-15	5	
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Two Channel SPST

### Two Channel SPST

DGM111	PMOS	75-200	20	0.3	1.0	0	0.5	4.6	10	-20	5	
DG133	N-JFET	30	20	0.6	1.6	1	0.8	2.5	12	-18	—	See DG181 For New Design
DG134	N-JFET	80	20	0.6	1.6	1	0.8	2.5	12	-18	—	See DG182 For New Design
DG141	N-JFET	10	20	1.0	2.5	1	0.8	2.5	12	-18	—	See DG180 For New Design
DG151	N-JFET	15	15	1.0	2.5	1	0.8	2.5	15	-15	—	See DG180 For New Design
DG152	N-JFET	50	15	0.8	1.6	1	0.8	2.5	15	-15	—	See DG181 For New Design
DG180	N-JFET	10	20	0.3	0.25	0	0.8	2.0	10	-20	5	Break-Before-Make
		10	15	0.3	0.25	0	0.8	2.0	15	-15	5	15V Supplies
<b>*DG181</b>	N-JFET	30	20	0.15	0.13	0	0.8	2.0	10	-20	5	Break-Before-Make
		30	15	0.15	0.13	0	0.8	2.0	15	-15	5	15V Supplies
<b>*DG182</b>	N-JFET	75	20	0.25	0.13	0	0.8	2.0	10	-20	5	Break-Before-Make
		75	20	0.25	0.13	0	0.8	2.0	15	-15	5	15V Supplies
<b>*DG200A</b>	CMOS Plus-40	70	30	1.0	0.5	0	0.8	2.0	15	-15	—	
DG281	N-JFET	300	20	0.15	0.13	0	0.8	2.0	15	-15	5	Break-Before-Make
<b>*DG300A</b>	CMOS	50	30	0.300	0.250	1	0.8	4.0	15	-15	—	
<b>*DG381A</b>	CMOS	50	30	0.300	0.250	0	0.8	4.0	15	-15	—	
<b>*DG304A</b>	CMOS	50	30	0.250	0.150	1	3.5	11.0	15	-15	—	CMOS compatible
<b>*DG5041</b>	CMOS Plus-40	50	30	1.0	0.5	1	0.8	2.0	15	-15	5	Break-Before-Make



Four Channel SPST

### Four Channel SPST

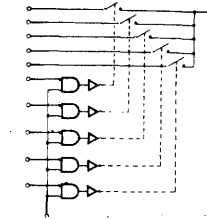
<b>*DG172</b>	PMOS	150-450	20	0.3	0.75	0	0.8	2.0	10	-20	5	
<b>*DG201A</b>	CMOS Plus-40	175	30	0.6	0.45	0	0.8	2.0	15	-15	—	
<b>*DG201HS</b>	CMOS Plus-40	50	30	.05	.04	0	0.8	2.0	15	-15	—	Super Fast Switch
<b>*DG221</b>	CMOS Plus-40	100	30			0	0.8	2.4	15	-15	—	with input latches
<b>*DG202</b>	CMOS Plus-40	175	30	0.6	0.45	1	0.8	2.0	15	-15	—	
<b>*DG211</b>	CMOS Plus-40	175	30	1.0	0.5	0	0.8	2.0	15	-15	5	
<b>*DG212</b>	CMOS Plus-40	175	30	1.0	0.5	1	0.8	2.0	15	-15	5	
<b>*DG308A</b>	CMOS Plus-40	100	30	0.2	0.15	1	3.5	11.0	15	-15	—	Single Supply Operation
<b>*DG309</b>	CMOS Plus-40	100	30	0.2	0.15	0	3.5	11.0	15	-15	—	Single Supply Operation

\*Devices recommended for new designs are indicated in bold face type.

# Analog Switch Product Information (Cont'd)

## Drivers and Gates (Cont'd)

Basic Part No. (Notes 1 & 2)	Switch Type	r <sub>DS(on)</sub> Max. (Ω) (Note 3)	Analog Voltage Range (p-pV) (Note 3)	Switching Time (μs)		Logic Input for ON Switch	Logic Levels (V)		Opt. Sup. Voltage (V)			Comments
				t <sub>ON</sub>	t <sub>OFF</sub>		V <sub>INL</sub>	V <sub>INH</sub>	(+) Sup. V <sub>1</sub>	(-) Sup. V <sub>2</sub>	Logic Sup. V <sub>I</sub>	



Five Channel SPST

### Five Channel SPST

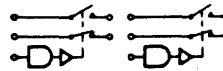
DG125	PMOS	100-450	20	0.3	2.0	0	0.5	4.6	10	-20	5	
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One Channel SPDT

### One Channel SPDT

DG143	N-JFET	80	20	0.8	1.6	(Note 4)	2.0	3.0	12	-18	—	See DG188 For New Design
DG144	N-JFET	30	20	0.8	1.6	(Note 4)	2.0	3.0	12	-18	—	See DG187 For New Design
DG146	N-JFET	10	20	1.0	2.5	(Note 4)	2.0	3.0	12	-18	—	See DG186 For New Design
DG161	N-JFET	15	15	1.0	2.5	(Note 4)	2.0	3.0	15	-15	—	See DG186 For New Design
DG162	N-JFET	50	15	0.8	1.6	(Note 4)	2.0	3.0	15	-15	—	See DG187 For New Design
DG186	N-JFET	10	20	0.3	0.25	(Note 5)	0.8	2.0	10	-20	5	Break-Before-Make
	N-JEFT	10	15	0.3	0.25	(Note 5)	0.8	2.0	15	-15	5	15V Supplies
*DG187	N-JFET	30	20	0.15	0.13	(Note 5)	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	30	15	0.15	0.13	(Note 5)	0.8	2.0	15	-15	5	15V Supplies
*DG188	N-JFET	75	20	0.25	0.13	(Note 5)	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	75	20	0.25	0.13	(Note 5)	0.8	2.0	15	-15	5	15V Supplies
DG287	N-JFET	300	20	0.15	0.13	(Note 5)	0.8	2.0	15	-15	5	Break-Before-Make
*DG301A	CMOS	50	30	0.300	0.250	(Note 5)	0.8	4.0	15	-15	—	
*DG387A	CMOS	50	30	0.300	0.250	(Note 5)	0.8	4.0	15	-15	—	
*DG305A	CMOS	50	30	0.250	0.150	(Note 5)	3.5	11.0	15	-15	—	CMOS compatible
SI3002	PMOS	100-400	20	1.0	1.5	(Note 5)	0.8	2.0	10	-20	—	
*DG5042	CMOS Plus-40	50	30	1.0	0.5	(Note 5)	0.8	2.0	15	-15	5	Break-Before-Make



Two Channel SPDT

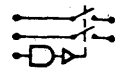
### Two Channel SPDT

DG189	N-JFET	10	20	0.3	0.25	(Note 5)	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	10	15	0.3	0.25	(Note 5)	0.8	2.0	15	-15	5	15V Supplies
*DG190	N-JFET	30	20	0.15	0.13	(Note 5)	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	30	15	0.15	0.13	(Note 5)	0.8	2.0	15	-15	5	15V Supplies
*DG191	N-JFET	75	20	0.25	0.13	(Note 5)	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	75	20	0.25	0.13	(Note 5)	0.8	2.0	15	-15	5	15V Supplies
*DG243	CMOS Plus-40	50	30	1.0	0.5	(Note 5)	0.8	2.0	15	-15	5	Make-Before-Break
DG290	N-JFET	300	20	0.15	0.13	(Note 5)	0.8	2.0	15	-15	5	Break-Before-Make
*DG303A	CMOS	50	30	0.300	0.250	(Note 5)	0.8	4.0	15	-15	—	
*DG390A	CMOS	50	30	0.300	0.250	(Note 5)	0.8	4.0	15	-15	—	
*DG307A	CMOS	50	30	0.250	0.150	(Note 5)	3.5	11.0	15	-15	—	CMOS compatible
*DG5043	CMOS Plus-40	50	30	1.0	0.5	(Note 5)	0.8	2.0	15	-15	5	Break-Before-Make

\*Devices recommended for new designs are indicated in bold face type.

# Analog Multiplexer Product Information

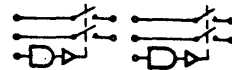
Basic Part No. (Notes 1 & 2)	Switch Type	r <sub>DS(on)</sub> Max. (Ω) (Note 3)	Analog Voltage Range (p-p V) (Note 3)	Switching Time (μs)		Logic Input for ON Switch	Logic Levels (V)		Opt. Sup. Voltage (V)			Comments
				t <sub>ON</sub>	t <sub>OFF</sub>		V <sub>INL</sub>	V <sub>INH</sub>	(+) Sup. V <sub>1</sub>	(-) Sup. V <sub>2</sub>	Logic Sup. V <sub>i</sub>	



One Channel DPST

## One Channel DPST

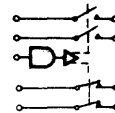
<b>*DG5044</b>	CMOS Plus-40	50	30	1.0	0.5	(Note 5)	0.8	2.0	15	-15	5	Break-Before-Make
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Two Channel DPST

## Two Channel DPST

DG126	N-JFET	80	20	0.6	1.6	(Note 9)	0.8	2.5	12	-18	—	See DG185 For New Design
DG129	N-JFET	30	20	0.6	1.6	(Note 9)	0.8	2.5	12	-18	—	See DG184 For New Design
DG140	N-JFET	10	20	1.0	2.5	(Note 9)	0.8	2.5	12	-18	—	See DG183 For New Design
DG153	N-JFET	15	15	1.0	2.5	(Note 9)	0.8	2.5	15	-15	—	See DG183 For New Design
DG154	N-JFET	50	15	0.6	1.6	(Note 9)	0.8	2.5	15	-15	—	See DG185 For New Design
DG183	N-JFET	10	20	0.3	0.25	1	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	10	15	0.3	0.25	1	0.8	2.0	15	-15	5	15V Supplies
<b>*DG184</b>	N-JFET	30	20	0.15	0.13	1	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	30	15	0.15	0.13	1	0.8	2.0	15	-15	5	15V Supplies
<b>*DG185</b>	N-JFET	75	20	0.25	0.13	1	0.8	2.0	10	-20	5	Break-Before-Make
	N-JFET	75	20	0.25	0.13	1	0.8	2.0	15	-15	5	15V Supplies
DG284	N-JFET	300	20	0.15	0.13	1	0.8	2.0	15	-15	5	Break-Before-Make
<b>*DG302 A</b>	CMOS	50	30	0.300	0.250	1	0.8	4.0	15	-15	—	
<b>*DG384 A</b>	CMOS	50	30	0.300	0.250	1	0.8	4.0	15	-15	—	
<b>*DG306 A</b>	CMOS	50	30	0.250	0.150	1	3.5	11.0	15	-15	—	CMOS compatible
<b>*DG5045</b>	CMOS Plus-40	50	30	1.0	0.5	1	0.8	2.0	15	-15	5	Break-Before-Make



One Channel DPDT

## One Channel DPDT

DG139	N-JFET	30	20	0.8	1.6	(Note 4)	2.0	3.0	12	-18	—	See DG191 For New Design
DG142	N-JFET	80	20	0.8	1.6	(Note 4)	2.0	3.0	12	-18	—	See DG190 For New Design
DG145	N-JFET	10	20	1.0	2.5	(Note 4)	2.0	3.0	12	-18	—	See DG189 For New Design
DG163	N-JFET	15	15	1.0	2.5	(Note 4)	2.0	3.0	15	-15	—	See DG189 For New Design
DG164	N-JFET	50	15	0.8	1.6	(Note 4)	2.0	3.0	15	-15	—	See DG191 For New Design

### NOTES:

- (1) \* Devices recommended for new designs are indicated in **bold face** type.
- (2) See individual data sheets for package and temperature designations.
- (3) Analog voltage range is a function of supply voltages. Where a FET switch is PMOS or CMOS, r<sub>DS</sub> is also a function of Supply Voltage and Analog Voltage. See individual data sheets for more detail.
- (4) Input reference voltage of 2.5 V is required (see data sheets).
- (5) See data sheet for switch state of differential switches.
- (6) Current Driven Device I<sub>INH</sub> = 1 mA.
- (7) For truth table see data sheet.
- (8) The appropriate switching characteristic for multiplexers is t<sub>transition</sub>, not t<sub>ON</sub>, t<sub>OFF</sub>.
- (9) Device normally operates with resistor to +10V.

# Drivers for MOSFET Switches

These drivers were designed to function as a level shifter and buffer between low level logic and the control gate of FET analog switches. Output voltage ratings are as high as 50V.

Basic Part Number (Note 2)		Function and Uses	at Rated Current(s)	OFF Level V(OUT)OFF at Rated Current or I(OUT)OFF at Rated Voltage	Input Logic for VOUT (low)	VINL (V)	VINH (V) (IINH) (mA)	Optimum Supply Voltage (V)				Switching Time, (μs)	
								V1	V2	VL	VR	tON	tOFF
D125	6 6	Six Separate MOSFET Drivers	0.4V@5mA	0.1μA@10V	0	0.5	4.6	(Note 9)	-20	5	-	0.5	1.2
D129	7 4	Four Channel (BV = 50) MOSFET Driver with Decode	0.7V@10mA	0.1μA@10V	1	0.7	2.2	(Note 9)	-20	-	-	0.25	0.8
*D169	2 4	Dual High-Speed Drivers with Complementary Outputs designed to drive high-capacity loads.	1.2V@1mA 3.0V@40mA	1.1V@1mA 2.5V@40mA	Output & Complement Available	0.8 0.8	2.0 2.0	15 15	-15 -15	5 5	0 0	t <sub>d</sub> <sup>+</sup> 0.17	t <sub>d</sub> <sup>-</sup> 0.20
*D469	4 4	Quad High Speed Driver	V <sub>OH</sub> V <sub>DD</sub> -2.0V @200mA V <sub>DD</sub> -0.2V @10mA	V <sub>OL</sub> 2.0V@200mA 0.2V@10mA	1	0.8	2.0	+12V, GND				0.045	
Si7250	4 8	Quad High Speed Driver	V <sub>DD</sub> -2.0V @200mA V <sub>DD</sub> -0.2V @10mA	2.0V@200mA 0.2V@10mA	Output & Complement Available	0.8	2.0	+12V, GND				0.150	

## Multiple FET Switches

Siliconix P-Channel MOSFET & DMOS Switches are available for such applications as sequential switching (commutation), signal processing, modulation, and A-to-D conversion. The MOSFET is normally OFF. These devices are also available with Siliconix drivers in a single package.

Basic Part Number (Note 2)	Circuit Function			Pull Up On Gate	r <sub>DS</sub> Max. (V)		BV <sub>DSS</sub>	I <sub>S(off)</sub> (nA)	V <sub>GS(th)</sub>		C <sub>gs</sub> Typ. (pF)	C <sub>ds</sub> Typ. (pF)	C <sub>sb</sub> Typ. (pF)
	S	D	G		@ V <sub>S</sub> = +10V	@ V <sub>S</sub> = -10V			Min.	Max.			
G115	6	1	6	SP6T	Yes	100	450	-30	0.5	-1.5 -4.0	0.9	0.4	2
G118	6	1	8	SP6T	No	100	450	-30	0.5	-1.5 -4.0	0.9	0.4	2
G119	6	2	3	DP3T	Yes	100	450	-30	0.5	-1.5 -4.0	1.8	0.4	2
G122	4	2	2	DPDT	Yes	100	450	-30	0.5	-1.5 -4.0	1.8	0.4	2
G123	4	2	4	2x SPDT	Yes	100	450	-30	0.5	-1.5 -4.0	1.8	0.4	2
*SD5000	4	4	4	4x SPST	No	50		20	10.0	0.1 2.0	3.5	0.5	4
*SD5001	4	4	4	4x SPST	No	50		10	10.0	0.1 2.0	3.5	0.5	4
*SD5002	4	4	4	4x SPST	No	50		15	10.0	0.1 2.0	3.5	0.5	4
*SD5200	4	4	4	4x SPST	No	80		30	1000	0.5 2.0	3.5	0.5	4

### NOTES:

- (1) \*Devices recommended for new designs are indicated in bold face type.
- (2) See individual data sheets for package and temperature designations.
- (9) Device normally operates with resistor to +10V.



## Pulse Width Modulators

Basic Part Number (Note 2)	Function and Uses	Operating Freq.	V <sub>OL</sub> @ Rated Current	V <sub>OH</sub> @ Rated Current	Supply Voltage
*PWM25	Pulse Width Modulator for SMPS to drive NPN or N channel MOSPOWER Devices	10Hz-400kHz with Dead Time Adjust	0.4V @ 20mA 2.5V @ 100mA	V+ -2V@20mA V+ -3V@100mA	+8.5V to +35V
*PWM27	Pulse Width Modulator for S SMPS to drive PNP or P channel MOSPOWER Devices	10Hz-400kHz with Dead Time	0.4V @ 20mA 2.5V @ 100mA	V+ -2V@20mA V+ -3V@100mA	+8.5V to +35V
*PWM125	Pulse Width Modulator for SMPS to drive N channel MOS-POWER Devices	10Hz - 800kHz with 100ns Dead Time	0.4V @ 20mA 2.5V @ 100mA	V+ -2V@20mA V+ -3V@100mA	+8.5V to +35V
*PWM127	Pulse Width Modulators for SMPS to drive P channel MOS-POWER Devices	10Hz - 800kHz with 100ns Dead	0.4V @ 20mA 2.5V @ 100mA	V+ -2V@20mA V+ -3V@100mA	+8.5V to +35V

## Voltage Converters

Basic Part Number (Note 2)	Function and Uses	Input Voltage Range	Quiescent Current	Output Voltage	Output Voltage @ Output Current
Si7661	Voltage Doubler/Inverter	+4.5V to +20V	2mA Max	-20V to +20V	-18V to +18V @ 20mA
Si7660	Voltage Converter/Inverter	+1.5V to +10.0V	See Data Sheet	-1.5V to -10.0V	See Data Sheet

### NOTES:

- (1) \*Devices recommended for new designs are indicated in **bold face** type.
- (2) See individual data sheets for package and temperature designations.
- (3) Analog voltage range is a function of supply voltages. Where a FET switch is PMOS or CMOS,  $r_{DS}$  is also a function of Supply Voltage and Analog Voltage. See individual data sheets for more detail.
- (4) Input reference voltage of 2.5 V is required (see data sheets).
- (5) See data sheet for switch state of differential switches.
- (6) Current Driven Device  $I_{INH} = 1$  mA.
- (7) For truth table see data sheet.
- (8) The appropriate switching characteristic for multiplexers is  $t_{transition}$ , not  $t_{ON}$ ,  $t_{OFF}$ .
- (9) Device normally operates with resistor to +10V.

# Analog Multiplexer Product Information

Basic Part No. (Notes 1 & 2)	Switch Type	r <sub>DS(on)</sub> Max. (Ω) (Note 3)	Analog Voltage Range (p-p V) (Note 3)	Switching Time (μs)		Logic Input for ON Switch	Logic Levels (V)		Opt. Sup. Voltage (V)			Comments
				t <sub>ON</sub>	t <sub>OFF</sub>		V <sub>INL</sub>	V <sub>INH</sub>	(+) Sup. V <sub>1</sub>	(-) Sup. V <sub>2</sub>	Logic Sup. V <sub>1</sub>	
<b>Eight Channel MUX + Enable</b>												
DG501	PMOS	150-250	10	1.5	(Note 8)	(Note 7)	0.6	3.5	5	-20	—	Logic Pullup Resistors
DG503	PMOS	150-800	20	1.5	(Note 8)	(Note 7)	0.6	8.5	10	-20	—	Note 9
<b>*DG508A</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.0	15	-15	—	
<b>*DG528</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.4	15	-15	—	With Input Latches
SI3705	PMOS	150-400	10	1.5	(Note 8)	(Note 7)	0.6	3.5	5	-20	—	See DG501/No Pullup Resistors
<b>Sixteen Channel MUX + Enable</b>												
<b>*DG506A</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.0	15	-15	—	Break-Before-Make
<b>*DG526</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.4	15	-15	—	Break-Before-Make
<b>Four Channel Differential MUX</b>												
<b>*DG509A</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.0	15	-15	—	Break-Before-Make
<b>*DG529</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.4	15	-15	—	With Input Latches
<b>Eight Channel Differential MUX + Enable</b>												
<b>*DG507A</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.0	15	-15	—	Break-Before-Make
<b>*DG527</b>	CMOS Plus-40	400	30	1.0	(Note 8)	(Note 7)	0.8	2.4	15	-15	—	With Input Latches
<b>Four Channel SPDT D/A Converter Summing Node Switches</b>												
DG515	NMOS	See Comments	—	0.120	0.170	(Note 5)	0.5	7.5	8.0	0	—	R <sub>1</sub> = 6.25Ω, R <sub>2</sub> = 12.5Ω, R <sub>3</sub> = 25Ω, R <sub>4</sub> = 50Ω
<b>Ten Channel SPDT D/A Converter Summing Node Switches</b>												
DG516	NMOS	See Comments	—	0.120	0.170	(Note 5)	0.5	7.5	8.0	0	—	R <sub>1</sub> = 100Ω, R <sub>2</sub> = 200Ω, R <sub>3</sub> = 400Ω, R <sub>4</sub> = 800Ω, R <sub>5</sub> = 1600Ω, R <sub>6-10</sub> = 3200Ω

**NOTES:**

- (1) \*Devices recommended for new designs are indicated in **bold face** type.
- (2) See individual data sheets for package and temperature designations.
- (3) Analog voltage range is a function of supply voltages. Where a FET switch is PMOS or CMOS, r<sub>DS</sub> is also a function of Supply Voltage and Analog Voltage. See individual data sheets for more detail.
- (4) Input reference voltage of 2.5 V is required (see data sheets).
- (5) See data sheet for switch state of differential switches.
- (6) Current Driven Device I<sub>INH</sub> = 1 mA.
- (7) For truth table see data sheet.
- (8) The appropriate switching characteristic for multiplexers is t<sub>transition</sub>, not t<sub>ON</sub>, t<sub>OFF</sub>.
- (9) Device normally operates with resistor to +10V.

# Linear and Data Conversion Product Information

## A/D and D/A Converters

3 1/2-Digit High Performance Integrating A/D Converter	<b>LD110/LD111A</b> 16-pin plastic DIPs	±3 1/2-Digit Resolution Accuracy 0.02% ±1 count Auto zero Auto polarity 10μV resolution Typical TC of 5 ppm/°C A usable 20mV scale	Three voltage ranges: 1.999V, 199.9mV & 19.99mV Sampling rate up to 40 samples/s Differential input capability Over-range & under-range signals TTL compatible
4 1/2-Digit Integrating A/D Converter	<b>LD120/121A</b> 16-pin & 18-pin plastic DIP respectively	±4 1/2-Digit Resolution Accuracy 0.005% ±1 count Auto zero Auto polarity TTL compatible Internal clock Linear to 28,500 counts	Two voltage ranges: 2.0V & 200.00mV 1 to 5 samples/s 25% inter-digit blanking MUX BCD outputs 0.5 count stability on 2.0V range Monolithic design
4 1/2-Digit Integrating A/D Converter	<b>LD122/121A</b> 16-pin & 18-pin plastic DIP respectively	±4 1/2-Digit Resolution Accuracy 0.005% ±1 count 1μV resolution for 20 mV FS Auto zero Auto polarity TTL compatibility Internal clock Linear to 28,500 counts	Same as LD120/121A but does not have internal input buffer amplifier
4 1/2-Digit Integrating A/D Converter	Si7135	±4 1/2-Digit Resolution Accuracy ±.005% ±1 count 2 Volts full scale Auto-zero TTL Compatible	Monolithic Construction Six Auxiliary I/O's are available for interfacing Micro-processors and UARTS 28 lead plastic pkg.
12 bit D/A Converter	Si7622	12 bit Multiplying D/A 12 bit Linearity Guaranteed Monotonic Single Supply Double Buffered Input Registers	Monolithic Construction 28 pin Plastic & Cerdip Micro-processor compatible
8 bit Data Acquisition System	Si520	8 bit by 8 Channel Data Acquisition System Ratiometric Conversion No Missing Codes Built-in 8-input Analog Multiplexer	28 pin Plastic Pkg. Latched 3 state outputs Latched Address Inputs Single 5V Supply

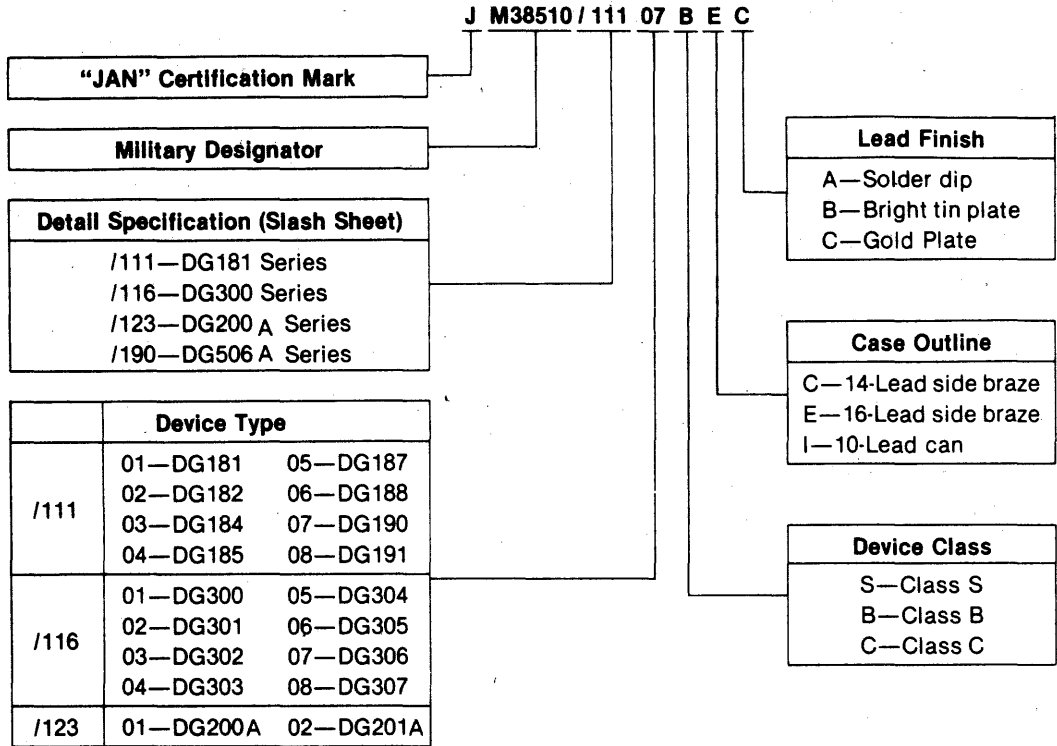
## Micropower Linear IC's

Triple Op Amp	<b>L144</b> 14-pin plastic, ceramic, flat-pack & Dice	±1.5 to ±18V supply Programmable supply current Internally compensated 0.4V/μs slew rate	80dB gain with 20kΩ load Drive large capacitive loads ±30V differential input Monolithic construction
Quad Comparator	<b>L161</b> 16-pin plastic, ceramic, flat-pack & Dice	±1.5 to ±18V supply Single supply operation Programmable supply current 3V/μs slew rate	Gain greater than 20V/mV Sensing near ground ±30V differential input CMOS Logic compatible

# Analog Switches and Multiplexers JAN38510

Several Siliconix Analog Switches and Multiplexers are available fully certified on the QPL (Qualified Parts List) published monthly by Defense Electronics Supply Center (DESC). The QPL numbers follow this format: JM38510/XXXXX. Refer to the current Siliconix Price List for available part types and order numbers.

## JAN Part Numbering System



Part Number	Order Part Number	Generic Part Number
JM38510/11101BCC	SJM181BCC	DG181AP/883
JM38510/11101BiC	SJM181BiC	DG181AA/883
JM38510/11102BCC	SJM182BCC	DG182AP/883
JM38510/11102BiC	SJM182BiC	DG182AA/883
JM38510/11103BEC	SJM183BEC	DG184AP/883
JM38510/11104BEC	SJM185BEC	DG185AP/883
JM38510/11105BCC	SJM187BCC	DG187AP/883
JM38510/11105BiC	SJM187BiC	DG187AA/883
JM38510/11106BCC	SJM188BCC	DG188AP/883
JM38510/11106BiC	SJM188BiC	DG188AA/883
JM38510/11107BEC	SJM190BEC	DG190AP/883
JM38510/11108BEC	SJM191BEC	DG191AP/883
JM38510/11601BCC	SJM300BCC	DG300AP/883
JM38510/11601BiC	SJM300BiC	DG300AA/883
JM38510/11602BCC	SJM301BCC	DG301AP/883
JM38510/11602BiC	SJM301BiC	DG301AA/883
JM38510/11603BCC	SJM302BCC	DG302AP/883
JM38510/11604BCC	SJM303BCC	DG303AP/883
JM38510/11605BCC	SJM304BCC	DG304AP/883
JM38510/11605BiC	SJM304BiC	DG304AA/883
JM38510/11606BCC	SJM305BCC	DG305AP/883
JM38510/11606BiC	SJM305BiC	DG305AA/883
JM38510/11607BCC	SJM306BCC	DG306AP/883
JM38510/11608BCC	SJM307BCC	DG307AP/883
JM38510/12303BCC	SJM200BCC	DG200A AP/883
JM38510/12303BiC	SJM200BiC	DG200A AA/883
JM38510/12304BEC	SJM201BEC	DG201A AP/883

## Fast CMOS Gate Arrays

The Siliconix family of CMOS gate arrays has been specifically designed to offer a fast and easy way to design your own custom integrated circuits. These FAST CMOS gate arrays offer virtually all the benefits of a full custom design at a fraction of the cost and time usually associated with full custom development programs.

Designing with our "State-of-the Art" silicon gate process is as easy as designing with standard "Off-the-Shelf" components. All the information necessary for you to begin your custom IC is contained in our design manual.

## No Previous IC Design Experience Necessary

Our design manual walks you through the complete design process. All the information necessary to take your conceptual design through the breadboard stage to the final prototype evaluation is discussed, with numerous examples given. Over 100 macrocells are included in the manual to assist in converting your logic diagrams to the gate level. Our applications engineers are always available to answer any questions you may have concerning the design and integration of your gate array. Included in the design manual are 200X layout sheets, macrocell planning sheets, test planning sheets, and other tools developed by IC design experts.

## A Wide Selection of Arrays to Choose From

Siliconix currently offers two families of oxide-isolated, silicon-gate arrays. Each family consists of seven different chips ranging in size from 360 to 1500 gates. A complete table of the arrays is shown in Table 1. The array family is determined by the process that is employed to fabricate the chips. The ISO-3 and ISO-5 refer to the design rules used for the gate masking step. Typical gate delays for each series are shown in Table 2 for  $V_{DD}$  of 5 volts.

Designation	Chip Size	Total Cells	Gate Equivalent	I/O Buffers	Bonding Pads
ISO5A/ISO3A	118x140	240	360	36	46
ISO5B/ISO3B	132x167	360	540	48	58
ISO5C/ISO3C	159x167	480	720	54	64
ISO5D/ISO3D	161x208	640	960	62	72
ISO5E/ISO3E	188x208	800	1200	68	78
ISO5F/ISO3F	188x244	1000	1500	76	86
ISO5G/ISO3G	214x244	1200	1800	84	92

TABLE 1

## Family of Arrays

Typical Gate Delays (ns) F.O. = 2

ISO-5				ISO-3			
GATE TYPE	Tr	Td	Tpd	GATE TYPE	Tr	Td	Tpd
INVERTER	4.3	1.7	3.0	INVERTER	1.35	0.45	0.9
2-INPUT NAND	4.5	3.5	4.0	2-INPUT NAND	1.5	1.0	1.25
2-INPUT NOR	9.5	2.1	5.8	2-INPUT NOR	2.85	0.55	1.7
3-INPUT NAND	4.8	6.0	5.4	3-INPUT NAND	1.6	1.7	1.65
3-INPUT NOR	13	2.5	7.8	3-INPUT NOR	4.4	0.6	2.5
4-INPUT NAND	5.0	7.0	6.0	4-INPUT NAND	1.7	2.3	2.0

TABLE 2

## Macrocells

Our design manual contains over 100 characterized, pre-designed macrocells for virtually all SSI functions. When these macrocells are used, all contact opening and metal routing within the cell is done automatically, eliminating the possibility of error when the function is implemented. The following example of the implementation of an exclusive-OR gate should help illustrate this point. The schematic is shown for reference only and is not necessary for the implementation of the function. The macrocell for an exclusive-OR gate is found in the design manual listing for exclusive-OR gates. Note that several different versions can be implemented. The macrocell chosen for this example is XO1A, as illustrated in Figure 3.

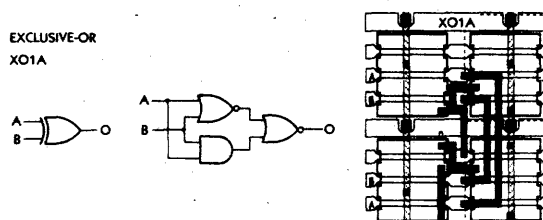


Figure 3






## Input/Output Buffers

The ISO series has very powerful I/O buffers designed to be direct TTL-compatible, capable of sinking 6mA at 0.4V or 24mA at 2.5V ( $V_{DD} = 5.0V$ ). Each p-buffer can source 2mA at 4.6V or 10mA at 2.5V. There are three transistors (one single and one double) in each buffer device which can be connected as desired for any particular application. All the output devices are fully guardbanded.

# Siliconix MOSPOWER Prime Products Selector Guide

\* 200°C RATING

Siliconix

Packages:								
BV <sub>DSS</sub> (Volts)	TO-3	TO-220	TO-39	TO-237	TO-92	TO-202	Quad Side Braze	Quad Plastic
<b>600-650</b>	VNT008A 6A, 1.5Ω, 650V	VNT008D 6A, 1.5Ω, 650V						
<b>450-500</b>	VNP006A 20A, 0.3Ω, 500V	IRF840 8A, 0.85Ω, 500V						
	IRF450 13A, 0.4Ω, 500V	VN5001D/IRF830 4.5A, 1.5Ω, 500V						
	IRF440 8A, 0.85Ω, 500V	IRF820 2.5A, 3Ω, 500V						
	VNP002A* 6.5A, 1.5Ω, 500V							
	VN5001A/IRF430 4.5A, 1.5Ω, 500V							
	IRF420 2.5A, 3Ω, 500V							
<b>350-400</b>	VNM005A 25A, 0.2Ω, 400V	IRF740 10A, 0.55Ω, 400V						
	IRF350 15A, 0.3Ω, 400V	VN4000D/IRF730 5.5A, 1.0Ω, 400V						
	IRF340 10A, 0.55Ω, 400V	IRF720 3A, 1.8Ω, 400V						
	VNM001A* 8A, 1.0Ω, 400V							
	VN4000A/IRF330 5.5A, 1.0Ω, 400V							
	IRF320 3.0A, 1.8Ω, 400V							
<b>120-240</b>	VNJ004A 45A, 0.06Ω, 200V	VN2406D 1.4A, 6Ω, 240V	VN2406B 0.8A, 6Ω, 240V	VN2406M 0.3A, 6Ω, 240V	VN2406L 0.21A, 6Ω, 240V			
	IRF250 30A, 0.085Ω, 200V	IRF640 18A, 0.18Ω, 200V		VN2410M 0.25A, 10Ω, 240V	VN2410L 0.16A, 10Ω, 240V			
	IRF240 18A, 0.18Ω, 200V	IRF630 9A, 0.4Ω, 200V			BS107 .1A, 28Ω, 200V			
	IRF230 9A, 0.4Ω, 200V	IRF620 5A, 0.8Ω, 200V						
	IRF220 5A, 0.8Ω, 200V							
<b>60-100</b>	VNE003A 60A, 0.035Ω, 100V	IRF540 27A, 0.085Ω, 100V	IRFF130 8A, 0.18Ω, 100V	VP1008M 0.37A, 5Ω, -100V	VP1008L 0.23A, 5Ω, -100V	VN88AF 1.5A, 4Ω, 80V	VQ1006P 0.40A, 4.5Ω, 90V	VQ1006J 0.40A, 4.5Ω, 90V
	IRF150 40A, 0.055Ω, 100V	VN1000D/IRF530 12A, 0.18Ω, 100V	IRFF120 6A, 0.30Ω, 100V	VN0808M 0.35A, 4Ω, 80V	VN0610L * 0.2A, 5Ω, 60V	VN80AF 1.3A, 5Ω, 80V	VQ2006P 0.41A, 5Ω, -90V	VQ2006J 0.41A, 5Ω, -90V
	IRF140 27A, 0.085Ω, 100V	IRF520 8A, 0.30Ω, 100V	VNE011B 4A, 0.5Ω, 100V	VN0606M 0.4A, 3Ω, 60V	VN0610LL 0.2A, 5Ω, 60V	VN66AF 1.7A, 3Ω, 60V	VQ1004P 0.46A, 3.5Ω, 60V	VQ1004J 0.46A, 3.5Ω, 60V
	VN1000A/IRF130 12A, 0.18Ω, 100V	VN88AD 1.7A, 4Ω, 80V	VP1008B 0.9A, 5Ω, -100V	VN10KM * 0.3A, 5Ω, 60V	BS170 .2A, 5Ω, 60V	VN67AF 1.6A, 3.5Ω, 60V	VQ1000P 0.225A, 5.5Ω, 60V	VQ1000J 0.225A, 5.5Ω, 60V
	IRF120 8A, 0.3Ω, 100V	VN0600D 16A, 0.12Ω, 60V	2N6661 ** 0.9A, 4Ω, 90V	VN10LM 0.3A, 5Ω, 60V	VN2222L * 0.15A, 7.5Ω, 60V		VQ2004P 0.41A, 5Ω, -60V	VQ2004J 0.41A, 5Ω, -60V
	VN0600A 16A, 0.12Ω, 60V	VN66AD 1.9A, 3Ω, 60V	2N6660 1.1A, 3Ω, 60V	VN2222LM 0.25A, 7.5Ω, 60V	VN2222LL 0.15A, 7.5Ω, 60V			
<b>30-50</b>	VN0400A 16A, 0.12Ω, 40V	VN0400D 16A, 0.12Ω, 40V	2N6659 1.4A, 1.8Ω, 35V	VN0300M 0.7A, 1.2Ω, 30V	BS250 1A, 14Ω, -45V	VN46AF 1.6A, 3Ω, 40V	VQ1001P 0.85A, 1.0Ω, 30V	VQ1001J 0.85A, 1.0Ω, 30V
	2N6656 2A, 1.8Ω, 35V	VN0300D 2.5A, 1.5Ω, 30V	VP0300B 1.3A, 2.5Ω, -30V	VP0300M 0.48A, 2.5Ω, -30V	VP0300L 3A, 2.5Ω, -30V	VN40AF 1.3A, 5Ω, 40V	VQ2001P 0.60A, 2Ω, -30V	VQ2001J 0.60A, 2Ω, -30V
<b>Specialty Products (N- and P-Channel Quad Arrays)</b>							VQ3001P ± 30V, 3Ω Total	VQ3001J ± 30V, 3Ω Total
							VQ7254P ± 20V, 3Ω Total	VQ7254J ± 20V, 3Ω Total

\* Zener Protected Gate \*\* JAN/JANTX/JANTXV QPL  
P-CHANNEL



# PERIPHERAL/DISPLAY DRIVER IC REFERENCE

INTERFACE

Sprague

Sink or Source in order of (1) output current rating, (2) output voltage rating, (3) number of drivers

TYPICAL DRIVING APPLICATIONS								OUTPUTS*			SPRAGUE PART NUMBER	EXTENDED TEMP. MILITARY HERM. AVAILABLE
LED	V. FLUOR.	DISPLAYS GAS DIS.	INCAND.	INDUCTIVE SOLENOID	MOTOR	PRINTERS THERMAL	ELECTRO. SENS.	mA	V	#		

## SINK DRIVERS

—	—	✓	—	—	—	—	—	20	-115	8	SERIES UDN-7180A	✓
✓	—	—	✓	—	—	✓	—	100	20	8†	UDN-2595A	✓
✓	—	—	✓	✓	✓	✓	—	250	100	4†	SERIES UHP-400/500	✓
SCR ARRAY	—	—	✓	✓	—	—	—	275	35	6†	UTN-2886B	—
SCR ARRAY	—	—	✓	✓	—	—	—	—	35	8†	UTN-2888A	—
✓	—	—	✓	✓	✓	✓	—	300	80	2†	SERIES UDN-3610M	✓
✓	—	—	✓	✓	✓	✓	—	—	80	2†	SERIES UDN-5710M	✓
✓	—	—	✓	✓	✓	✓	—	—	80	4†	SERIES UDN-5700A	✓
PIN DIODE DRIVER	—	—	✓	—	—	—	—	120	4	—	UDN-5791A	✓
✓	—	—	✓	✓	✓	✓	✓	350	50	4†	UCN-4401A	✓
✓	—	—	✓	✓	✓	✓	✓	—	50	7	SERIES ULN-2000A	✓
✓	—	—	✓	✓	✓	✓	✓	—	50	9	SERIES ULN-2800A	✓
✓	—	—	✓	✓	✓	✓	✓	—	50	8†	UCN-4801A	✓
✓	—	—	✓	✓	✓	✓	✓	—	50	8†	UCN-4821A	✓
✓	—	—	✓	✓	✓	✓	✓	—	70	2†	SERIES UDN-5720M	✓
✓	—	—	✓	✓	✓	✓	✓	—	80	8†	UCN-4822A	✓
✓	—	—	✓	✓	✓	✓	✓	—	95	7	SERIES ULN-2020A	✓
✓	—	—	✓	✓	✓	✓	✓	—	95	8	SERIES ULN-2820A	✓
✓	—	—	✓	✓	✓	✓	✓	—	100	8§	UCN-4823A	✓
STEPPER MOTOR DRIVER	—	—	✓	—	✓	—	—	500	50	4	UCN-4202/03A	✓
✓	—	—	✓	✓	✓	✓	—	—	50	7	SERIES ULN-2010A	✓
✓	—	—	✓	✓	✓	✓	—	—	50	8	SERIES ULN-2810A	✓
✓	—	—	✓	✓	✓	✓	—	—	50	8†	UCN-4807/08A	✓
✓	—	—	✓	✓	✓	✓	—	600	70	2†	SERIES UDN-5730/40M	✓
✓	—	—	✓	✓	✓	—	—	1000	80	4†	UDN-2542B	—
✓	—	—	✓	✓	✓	—	—	1250	50	2	ULN-2061M	—
✓	—	—	✓	✓	✓	—	—	—	50	4	SERIES ULN-2064B	✓
✓	—	—	✓	✓	✓	—	—	—	60	4†	UDN-2541B	—
✓	—	—	✓	✓	✓	—	✓	1500	-50	4	UDN-2841/45B	—
✓	—	—	✓	✓	✓	—	—	—	80	2	ULN-2062M	—
✓	—	—	✓	✓	✓	—	—	—	80	4	SERIES ULN-2065B	✓
HALF-BRIDGE	—	—	—	—	✓	—	—	±2000	30	1	UDN-2949Z	—
HALF-BRIDGE	—	—	—	—	✓	—	—	—	35	1	UDN-2935/50Z	—
FULL-BRIDGE	—	—	—	—	✓	—	—	—	40	1	UDN-2952B/W	—
✓	—	—	✓	✓	✓	—	—	±4000	60	2	UDN-2975/76W	—
✓	—	—	✓	✓	✓	—	—	4000	80	4	UDN-2878/79W	—

## SOURCE DRIVERS

—	✓	—	—	—	—	—	—	-25	±40	8	UDN-6138/48A	—
✓	✓	—	—	—	—	DECODER/DRIVERS	—	—	60	8†	UCN-4805A	—
✓	✓	—	—	—	—	✓	—	—	60	20§	UCN-5812A	✓
✓	✓	—	—	—	—	✓	—	—	60	32§	UCN-5818A	✓
—	✓	—	—	—	—	✓	—	—	80	8†	UCN-4815A	✓
✓	✓	—	—	—	—	✓	—	—	80	10§	UCN-4810A	✓
✓	✓	—	—	—	—	✓	—	—	80	10§	UCN-5810A	✓
—	✓	—	—	—	—	—	—	—	115	6	UDN-6116A	✓
—	✓	—	—	—	—	—	—	—	115	8	UDN-6118A	✓
—	✓	—	—	—	—	—	—	—	140	8	UDN-6514A	—
—	✓	—	—	—	—	—	—	—	200	8	UDN-6510A	—
✓	—	—	✓	✓	✓	✓	—	-120	25	8†	UDN-2585A	✓
✓	—	—	✓	✓	✓	✓	✓	-350	50	8	UDN-2981/82A	✓
✓	—	—	✓	✓	✓	✓	✓	—	50	8§	UCN-5890A	✓
✓	—	—	✓	✓	✓	✓	✓	—	80	8	UDN-2983/84A	✓
—	✓	—	—	—	—	—	—	—	-80	5	UDN-2956/57A	—
✓	—	—	✓	✓	✓	—	—	—	-80	8	UDN-2580A	✓
✓	—	—	✓	✓	✓	—	—	—	-80	8	UDN-2588A	✓
✓	—	—	✓	✓	✓	—	—	-1500	30	4	UDN-2941B/W	—
✓	—	—	✓	✓	✓	—	—	—	-50	4	UDN-2845B	—
HALF-BRIDGE	—	—	—	—	✓	—	—	±2000	30	1	UDN-2949Z	—
HALF-BRIDGE	—	—	—	—	✓	—	—	—	35	1	UDN-2935/50Z	—
FULL-BRIDGE	—	—	—	—	✓	—	—	—	40	1	UDN-2952B/W	—
—	—	—	—	✓	—	—	—	±4000	60	2	UDN-2975/76W	—

\*Current is maximum tested condition, voltage is absolute maximum rating. †Latched Drivers. §Serial-input, latched parallel-outputs. ††Saturated, non-Darlington outputs for low-voltage applications.

For application engineering assistance, write or call Semiconductor Division, Sprague Electric Company, 115 Northeast Cutoff, Worcester, Mass. 01606. Tel. 617/853-5000.

For the name of your nearest Sprague Semiconductor Distributor, write or call Sprague Products Division, Sprague Electric Company, North Adams, Mass. 01247. Tel. 413/664-4481.

4SS-2159R1

**SPRAGUE**

THE MARK OF RELIABILITY

a Penn Central unit

**QUAD HIGH-CURRENT  
DARLINGTON SWITCH ICs****UDN-2878W AND UDN-2879W  
QUAD HIGH-CURRENT DARLINGTON SWITCHES****FEATURES**

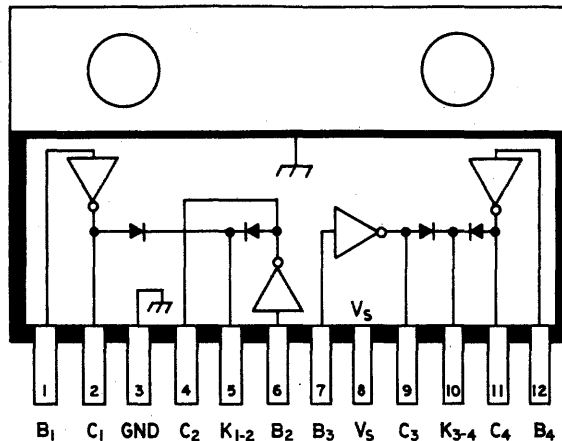
- Output Currents to 4 A
- Output Voltages to 80 V
- Loads to 1280 W
- TTL, DTL, or CMOS Compatible Inputs
- Internal Clamp Diodes
- Plastic Single In-Line Package
- Heat-Sink Tab

THESE QUAD DARLINGTON ARRAYS are designed to serve as interface between low-level logic and peripheral power devices such as solenoids, motors, incandescent displays, heaters, and similar loads of up to 320 W per channel. Both integrated circuits include transient-suppression diodes that enable use with inductive loads. The input logic is compatible with most TTL, DTL, LS TTL, and 5 V CMOS logic.

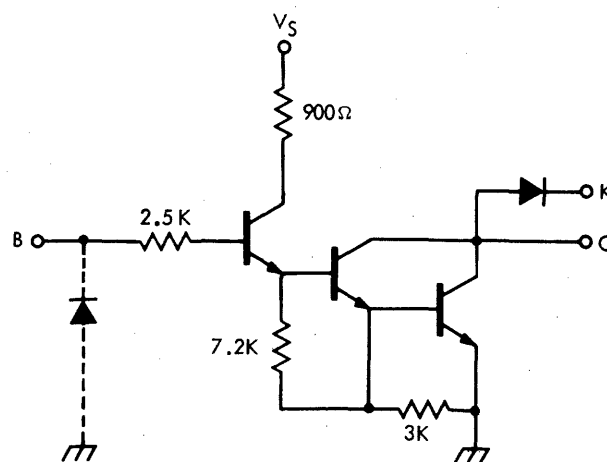
For maximum power-handling capability, all drivers are supplied in a 12-pin single in-line power-tab package. The tab is at ground potential and needs no insulation. External heat sinks are usually required for proper operation of these devices.

Similar integrated circuits (Types UDN-2068B and UDN-2069B) are available in 16-pin dual in-line packages for operation with load currents of up to 1.5 A per channel. They are described in the most recent issue of Sprague Engineering Bulletin 29305.

Device	Sustaining Voltage	Output Voltage	Output Current
UDN-2878W	35 V	50 V	4 A
UDN-2878W-2	35 V	50 V	3 A
UDN-2879W	50 V	80 V	4 A
UDN-2879W-2	50 V	80 V	3 A



DWG. NO. A-11,974

**PARTIAL SCHEMATIC  
One of 4 Drivers**

DWG. NO. A-12,037

Write for Engineering Bulletin 29305.10 to Technical Literature Service, Sprague Electric Company, 647 Marshall St., North Adams, Mass. 01247

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# FULL-BRIDGE MOTOR DRIVER INTEGRATED CIRCUITS

INTERFACE  
Sprague

## UDN-2952B AND UDN-2952W FULL-BRIDGE MOTOR DRIVERS

### FEATURES

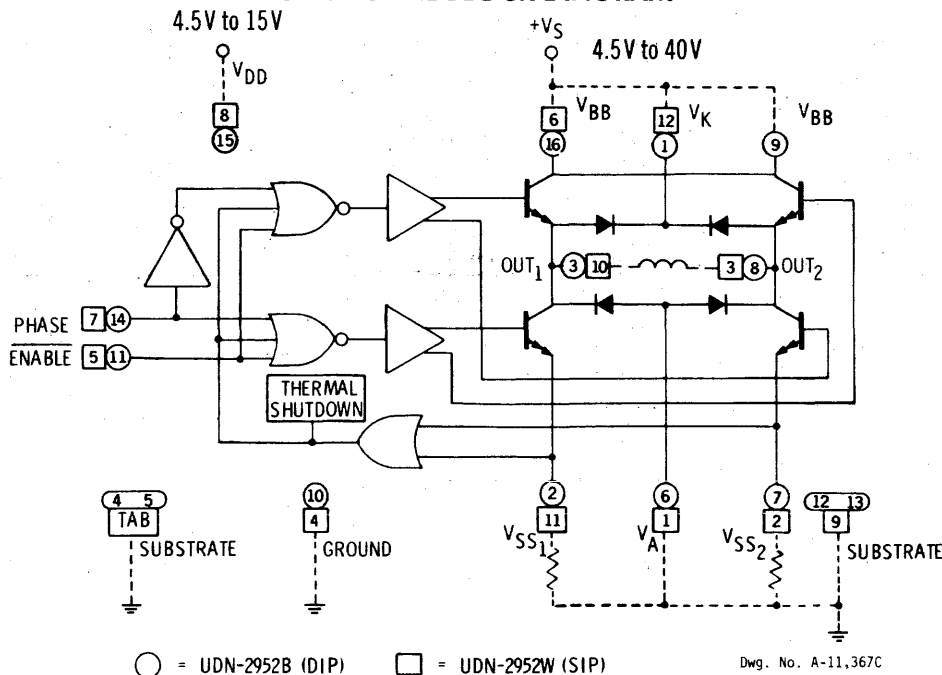
- Output Currents to  $\pm 3.5A$
- Output Voltages to 40V
- Adjustable Short-Circuit Protection
- Thermal Protection
- Internal Clamp Diodes
- TTL, DTL, PMOS, CMOS Compatible
- DIP or SIP Packaging

**FULL-BRIDGE MOTOR-DRIVER** integrated circuits, Types UDN-2952B and UDN-2952W combine low-level logic circuitry and Darlington output power drivers for bidirectional control of d-c motors or solenoids operating with continuous load currents of up to 2A and peak start-up currents as high as 3.5 A.

These monolithic integrated circuits have extensive circuit protection. Both drivers have thermal shutdown networks that disable motor drive if the package power dissipation ratings are exceeded. Internal diode transient suppression is provided on-chip. Output-current limiting is determined by the user's selection of a sensing resistor.

The Type UDN-2952B full-bridge power driver is supplied in a 16-pin dual in-line plastic package with copper heat-sink contact tabs. The lead configuration enables easy attachment of a heat sink while fitting a standard integrated circuit socket or printed wiring board layout. Type UDN-2952W, for higher power requirements, is in a 12-pin single in-line power tab package. The tab is at ground potential and needs no insulation.

### FUNCTIONAL BLOCK DIAGRAM



Write for Engineering Bulletin 29319 to Technical Literature Service, Sprague Electric Company, 647 Marshall Street, North Adams, Mass. 01247

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455-3160

# ABBREVIATIONS OF COMPANY NAMES

<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.	<b>Harris</b>	Harris Semiconductor	<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Heurikon</b>	Heurikon Corp.	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hitachi</b>	Hitachi America, Ltd.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Holt</b>	Holt Inc.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>HP</b>	Hewlett-Packard		
<b>Analogic</b>	Analogic	<b>Hughes</b>	Hughes Aircraft, Solid State Products		
<b>Analog Sys</b>	Analog Systems				
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Panasonic</b>	Panasonic
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>Pico Design</b>	Pico Design
<b>APM</b>	Applied Microsystems Corp.			<b>Polycore</b>	Polycore Electronics
<b>Appl Sys</b>	Applied Systems Corp.	<b>ICC</b>	International Cybernetics	<b>Plessey</b>	Plessey Semiconductors
<b>APT</b>	Applied Microtechnology	<b>IDT</b>	Integrated Device Technology	<b>PMI</b>	Precision Monolithics, Inc.
<b>Aptek</b>	Aptek Microsystems	<b>IMI</b>	International Microcircuits, Inc.	<b>PragDes</b>	Pragmatic Design Inc.
<b>Array Tech</b>	Array Technology	<b>IMP</b>	International Microelectronic Products	<b>Pro-Log</b>	Pro-Log Corp.
<b>AWI</b>	AWI Electronics				
		<b>IMS</b>	Industrial MicroSystems Inc.	<b>Quay</b>	Quay Corp.
<b>Barvon</b>	Barvon Research	<b>Infosphere</b>	Infosphere		
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>Inmos</b>	Inmos	<b>Raytheon</b>	Raytheon Semiconductor
<b>Burr-Brown</b>	Burr-Brown	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>RCA</b>	RCA Solid State Division
		<b>IntCirSys</b>	Integrated Circuit Systems	<b>RCI Data</b>	RCI Data
<b>CAE</b>	Computer Aided Engineering	<b>IntCompSys</b>	Integrated Computer Systems	<b>RELMS</b>	Relational Memory Systems
<b>Cal Devices</b>	California Devices	<b>Int Tech</b>	Integrated Technology Corp.	<b>Reticon</b>	Reticon
<b>Cermetek</b>	Cermetek	<b>Intech</b>	Intech Microcircuits Div.	<b>RIFA</b>	RIFA
<b>CGRS</b>	CGRS Microtech Inc.	<b>Intel</b>	Intel	<b>Rockwell</b>	Rockwell, Microelectronic Devices
<b>Cherry</b>	Cherry Semiconductor	<b>Interdesign</b>	Interdesign	<b>RTC</b>	Riehl Time Corporation
<b>CIC</b>	Custom Integrated Circuits	<b>Intersil</b>	Intersil		
<b>CirTech</b>	Circuit Technology	<b>Intronics</b>	Intronics	<b>Sanyo</b>	Sanyo
<b>Citel</b>	Citel	<b>ITT</b>	ITT Semiconductors	<b>SBE</b>	SBE, Inc.
<b>Comlinear</b>	Comlinear Corporation			<b>SEQ</b>	SEQ Technology, Inc.
<b>CMA</b>	Custom MOS Arrays	<b>Kinetic Sys</b>	Kinetic Systems	<b>SPI</b>	Semi Processes Inc.
<b>Comark</b>	Comark Corp.	<b>Kontron</b>	Kontron Electronics	<b>Siemens</b>	Siemens
<b>Comdial</b>	Comdial Semiconductor			<b>Si-Fab</b>	Si-Fab
<b>Comp Auto</b>	Computer Automation	<b>Lambda</b>	Lambda Semiconductor	<b>Signetics</b>	Signetics
<b>Compas</b>	Compas Microsystems	<b>Linear Tech</b>	Linear Technology	<b>SGS</b>	SGS Semiconductor
<b>Cont Logic</b>	Control Logic Inc.	<b>LSI Comp</b>	LSI Computer Systems	<b>Sharp</b>	Sharp
<b>Control Sys</b>	Control Systems Microsystems Div.	<b>LSI Logic</b>	LSI Logic Corporation	<b>Silicon G</b>	Silicon General
<b>CreMicro</b>	Creative Micro Systems			<b>Siliconix</b>	Siliconix
<b>Cromemco</b>	Cromemco, Inc.	<b>Master Logic</b>	Master Logic Corporation	<b>Silicon Sys</b>	Silicon Systems Inc.
<b>Cubit</b>	Cubit Inc.	<b>Matrix</b>	Matrix Corp.	<b>Siltronics</b>	Siltronics
<b>Curtis</b>	Curtis Electro Devices, Inc.	<b>Matrox</b>	Matrox Electronic Systems	<b>SMC</b>	Standard Microsystems Corp.
<b>Cybernetic</b>	Cybernetic Micro Systems	<b>Maxim</b>	Maxim Integrated Products	<b>S MOS</b>	S MOS Systems
<b>Cybersys</b>	Cybersystems	<b>MCC</b>	Micro-Computer Control	<b>Solarise</b>	Solarise Enterprises
<b>Cybertek</b>	Cybertek Inc.	<b>MCE</b>	MCE Electronics	<b>Solitron</b>	Solitron Devices
		<b>Micrel</b>	Micrel	<b>Sprague</b>	Sprague Electric Company
<b>Data General</b>	Data General	<b>Micro Innov</b>	Micro Innovators	<b>SSM</b>	Solid State Micro Technology for Music
<b>Data I/O</b>	Data I/O	<b>Micropac</b>	Micropac Industries	<b>SSS</b>	Solid State Scientific
<b>Data Trans</b>	Data Translation	<b>Micro Net</b>	Micro Networks	<b>Stag</b>	Stag Microsystems
<b>Datel</b>	Datel	<b>Micro Pwr</b>	Micro Power Systems	<b>STC</b>	Storage Technology Corp.
<b>Datricon</b>	Datricon Corporation	<b>Micro Sci</b>	Micro Sciences Corp.	<b>STD</b>	STD Microsystems
<b>DDC</b>	Data Devices Corporation	<b>Micro Tech</b>	Microcircuits Technology	<b>Struc Des</b>	Structured Design Inc.
<b>DEC</b>	Digital Equipment Corporation	<b>Micro-Link</b>	Micro-Link Corporation	<b>Stynetic</b>	Stynetic Systems
<b>Die-Tech</b>	Die-Tech	<b>Micron</b>	Micron Technology	<b>Sunrise</b>	Sunrise Electronics
<b>Digelec</b>	Digelec Corp.	<b>MillerTron</b>	MillerTronics	<b>Sunshine</b>	Sunshine Semiconductor
<b>Digitek</b>	Digitek, Inc.	<b>Miller</b>	Miller Technology	<b>Supertex</b>	Supertex Inc.
<b>Dionics</b>	Dionics Inc.	<b>Mitel</b>	Mitel Semiconductor	<b>Symtek</b>	Symtek Corp.
<b>Dist Comp</b>	Distributed Computer Systems	<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Synertek</b>	Synertek
<b>Divers Tech</b>	Diversified Technology	<b>MNI</b>	Monolithic Memories, Inc.	<b>Sys Innov</b>	Systems Innovations
		<b>Mostek</b>	Monolithic Systems Corp.		
<b>E-HI</b>	E-H International, Inc.	<b>Motorola</b>	Motorola Semiconductor	<b>Tau Zero</b>	Tau Zero Inc.
<b>EDI</b>	Electronic Designs Inc.	<b>MRC</b>	MRC Systems	<b>Technitrol</b>	Technitrol
<b>Elind</b>	Elind Elettronica Industriale	<b>Murray</b>	Murray Consulting	<b>Tektronix</b>	Tektronix
<b>EMM-SESCO</b>	EEM-SESCO	<b>Monosil</b>		<b>Teledyne C</b>	Teledyne Crystalonics
<b>Emulogic</b>	Emulogic Inc.			<b>Teledyne P</b>	Teledyne Philbrick
<b>ETI Micro</b>	ETI Micro	<b>National</b>	National Semiconductor	<b>Teledyne S</b>	Teledyne Semiconductor
<b>Exar</b>	Exar Integrated Systems	<b>NCM</b>	NCM Corp.	<b>Telefunken</b>	Telefunken
<b>Exel</b>	Exel Microelectronics	<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Telmos</b>	Telmos
		<b>NEC</b>	NEC Electronics	<b>Teltone</b>	Teltone Corporation
<b>Fairchild</b>	Fairchild	<b>Nitron</b>	Nitron	<b>TI</b>	Texas Instruments
<b>Ferranti</b>	Ferranti Electric			<b>Third Domain</b>	Third Domain
<b>Force</b>	Force Computers	<b>National</b>	National Semiconductor	<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
<b>Fujitsu A</b>	Fujitsu America	<b>NCM</b>	NCM Corp.	<b>Toshiba</b>	Toshiba America
<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.	<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Trans-Data</b>	Trans-Data
				<b>TRW</b>	TRW LSI Products
				<b>Unitrode</b>	Unitrode
				<b>Universal</b>	Universal Semiconductor, Inc.
				<b>Varix</b>	Varix Corp.
				<b>VLSI Design</b>	VLSI Design Associates
				<b>VTI</b>	VLSI Technology, Inc.
				<b>Votrax</b>	Votrax
				<b>Weitek</b>	Weitek Corporation
				<b>Western</b>	Western Digital
				<b>Wintek</b>	Wintek Corp.
				<b>Xicor</b>	Xicor, Inc.
				<b>Xycom</b>	Xycom
				<b>Zendex</b>	Zendex Corp.
				<b>Zilog</b>	Zilog
				<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrex</b>	Zytrex Corp.

**General Description**

The TSC426/427/428 are dual high speed power MOSFET drivers. Each device converts CMOS/TTL input levels to a high voltage level drive signal. Switching times of 50 ns are typical with 1000 pf loads. Peak output drive current is 1.5 amp. Delay times have been matched for both inverting and non-inverting stages.

The TSC428 contains an inverting and non-inverting driver. The TSC426 is inverting and TSC427 non-inverting.

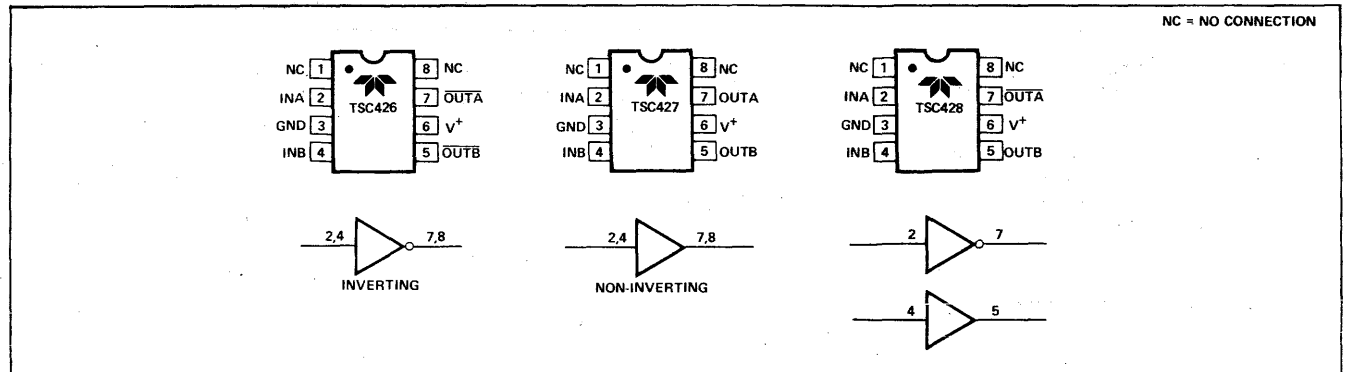
**Features**

- CMOS Construction for Low Current Drain ..... 6 mA
- High Output Voltage .....  $V_s - 0.05 V$
- High Peak Current Drive ..... 1.5 A
- High Speed Switching ( $C_H = 1000 pF$ ) ..... 50 ns
- Available as Inverting/Non-Inverting Driver
- Wide Supply Voltage Operation ..... 4.5 to 20 V
- TTL/CMOS Compatible Input Drive
- Pin-Out Equivalent to DS0026

**Preliminary Ordering Information**

Ordering Part No.	Operating Temp. Range	Package	Configuration	Peak Output Current	Rise Time ( $C_L = 1000 pF$ )
TSC426CPA	0°C to 70°C	8-Pin Plastic Dip	Inverting	1.5 A	50 ns
TSC426IJA	-25°C to 85°C	8-Pin CerDIP	Inverting	1.5 A	50 ns
TSC426MJA	-55°C to 125°C	8-Pin CerDIP	Inverting	1.5 A	50 ns
TSC427CPA	0°C to 70°C	8-Pin Plastic Dip	Non-Inverting	1.5 A	50 ns
TSC427IJA	-25°C to 85°C	8-Pin CerDIP	Non-Inverting	1.5 A	50 ns
TSC427MJA	-55°C to 125°C	8-Pin CerDIP	Non-Inverting	1.5 A	50 ns
TSC428CPA	0°C to 70°C	8-Pin Plastic Dip	Non-Inv. & Inv.	1.5 A	50 ns
TSC428IJA	-25°C to 85°C	8-Pin CerDIP	Non-Inv. & Inv.	1.5 A	50 ns
TSC428MJA	-55°C to 125°C	8-Pin CerDIP	Non-Inv. & Inv.	1.5 A	50 ns

**Pin Configuration**



### General Description

The TSC700A drives common anode LED displays with 28 high current, open-drain N channel output transistors. Four seven segment LED displays may be driven. Drive current is guaranteed to be 11 mA minimum. This is twice the minimum drive current available from comparable devices and will provide high LED luminance. High luminous intensity is an important factor when a dark contrasting background is unavailable or the LED is viewed at a distance. The TSC700A current capability makes it an ideal large character LED driver.

Four data bit inputs and four digit select signals permit interfacing to multiplexed BCD or binary output devices. The four bit data input is decoded into the seven segment alphanumeric code known as "Code B". A 0 to 9, —, E, H, L, P or "blank" reading may be displayed.

An added feature includes a brightness control input that adjusts segment drive current. The control pin may also be used as a digital display enable. The TSC700A is an improved pin compatible and functional equivalent to the ICM7212A and TSC7212A.

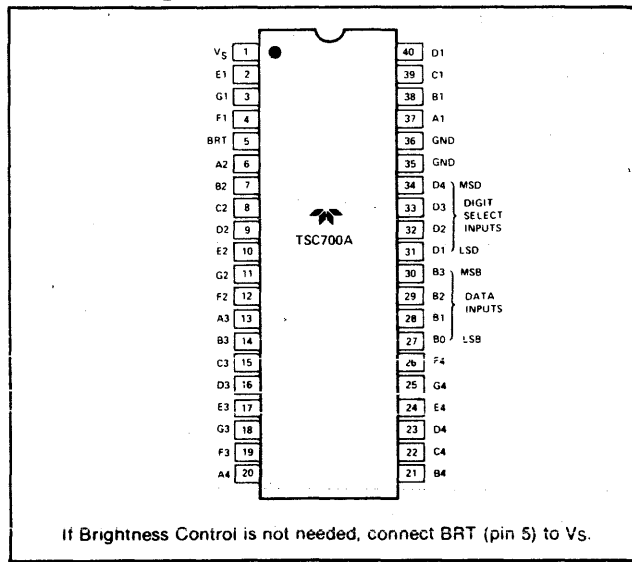
### Ordering Information

Part No.	Package	Temp. Range	LED Segment Current	Output Code
TSC700AMJL	40 Pin CerDIP	-55°C to +125°C	14 mA	Code B
TSC700AIJL	40 Pin CerDIP	-25°C to +85°C	14 mA	Code B
TSC700AMJL/883	40 Pin CerDIP	-55°C to +125°C	14 mA	Code B
TSC700A/Y	CHIP	25°C	14 mA	Code B

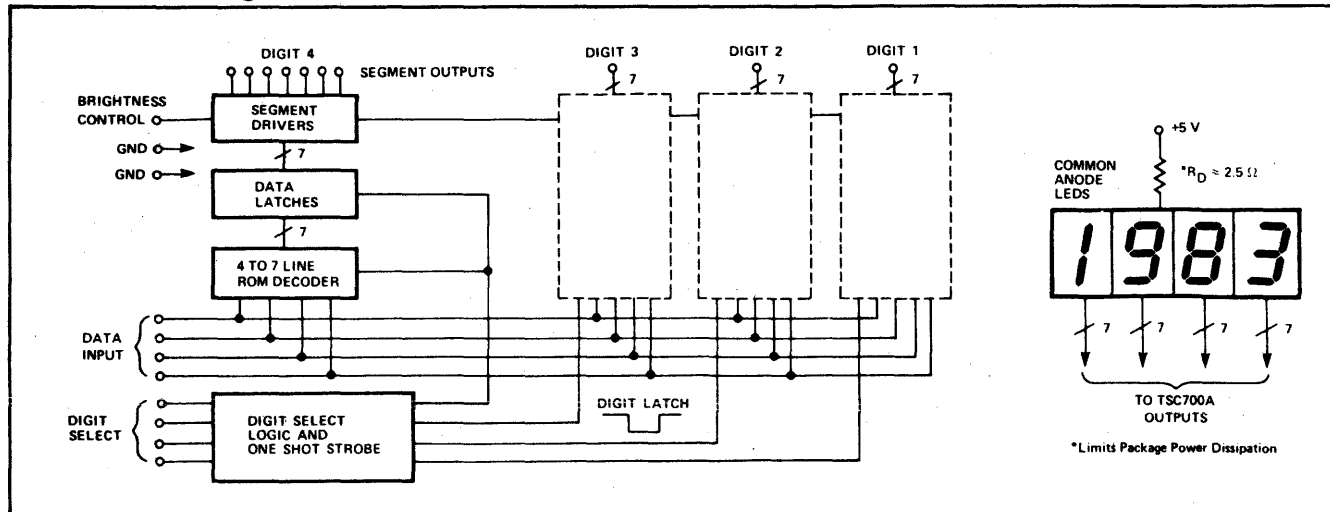
### Features

- High Drive Current for High Luminance LED Display
- Guaranteed High LED Segment Current 11 mA Minimum
- 28 Common Anode LED Drivers (4 Digits)
- Code B Output Format . . . 0 to 9, —, E, H, L, P, "blank"
- BCD/Binary Input to Seven Segment LED Code
- Four Separate Digit Selects for Multiplexed Input
- Digital or Analog Brightness Control
- Digital Display Enable
- Low Thermal Resistance Package
- Military Temperature Range Devices Available
- Pin Compatible With TSC7212A, ICM7212A

### Pin Configuration



### Functional Diagram



**General Description**

The TSC701AM is a CMOS direct drive, four digit, seven segment LED display decoder and driver. The device is bus compatible making microprocessor controlled displays possible. Two chip select signals control data and digit select code latching prior to decoding and display. External data latches are unnecessary.

The TSC701AM drives common anode LED displays with 28 high current, open-drain N channel output transistors. Four seven segment LED displays may be driven. Drive current is guaranteed to be 11 mA minimum (18 mA TYP). This is twice the minimum drive current available from comparable devices and will provide high LED luminance. High luminous intensity is an important factor when a dark contrasting background is unavailable or the LED is viewed at a distance. The TSC701AM current capability makes it an ideal large character LED driver.

Four data bit inputs and four digit select signals permit interfacing to multiplexed BCD or binary output devices. The four bit data input is decoded into the seven segment alphanumeric code known as "Code B". A 0 to 9, —, E, H, L, P or "blank" reading may be displayed.

An added feature is the brightness control input that adjusts segment drive current. The control pin may also be used as a digital display enable. The TSC701AM is an improved pin compatible and functional equivalent to the ICM7212AM.

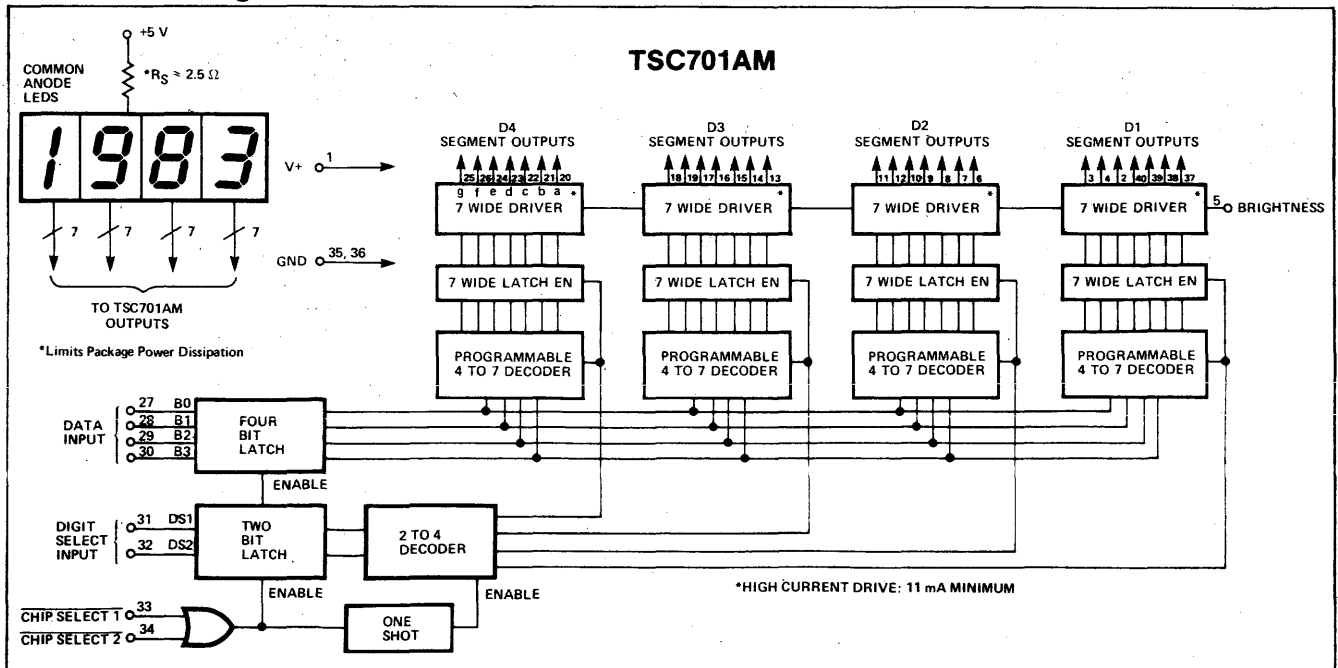
**Features**

- 28 Current Limited Outputs Drive Common-Anode LEDs at 18 mA Per Segment.
- Input and Digit Select Data Latches.
- Brightness Input Allows Potentiometer Control of LED Segment Current. Pin Also Serves as Digital Display Enable.
- Input and Digit Select Data Latches.
- Pin Compatible and Functionally Equivalent to ICM7212AM.
- Input Decoded to Seven Segment Code B Output (0 to 9, —, E, H, L, P, "Blank")

**Ordering Information**

Part No.	Package	Temp. Range	LED Segment Current	Output Code
TSC701AMIJL	40-Pin CerDIP	-25°C to +85°C	18 mA	Code B

**Functional Diagram**



### General Description

The TSC7109 is a 12-bit plus sign CMOS low power A/D Converter. The single CMOS IC contains all the necessary active devices to interface with microprocessors.

In direct mode, Chip Select and High/Low Byte Enables control parallel bus interface. In the handshake mode the TSC7109 will operate with industry standard UART's in controlling serial data transmission, ideal for remote data logging. Control and monitoring of conversion timing is provided by the RUN/HOLD and STATUS outputs. The TSC7109 requires only the addition of eight passive components plus a crystal to operate as a dual slope integrating A/D converter. The TSC7109 has features that make it an attractive per-channel alternative to analog multiplexing for many data acquisition applications. These features include typical input bias current of 1 pA, drift of less than 1  $\mu\text{V}/^\circ\text{C}$ , input noise typically 15  $\mu\text{V}$  p-p, and auto-zero. True differential input and reference allows the measurement of bridge-type transducers such as load cells, strain gauges and temperature transducers.

For applications requiring more resolution see the TSC800, 15-bit plus sign data sheet.

### Ordering Information

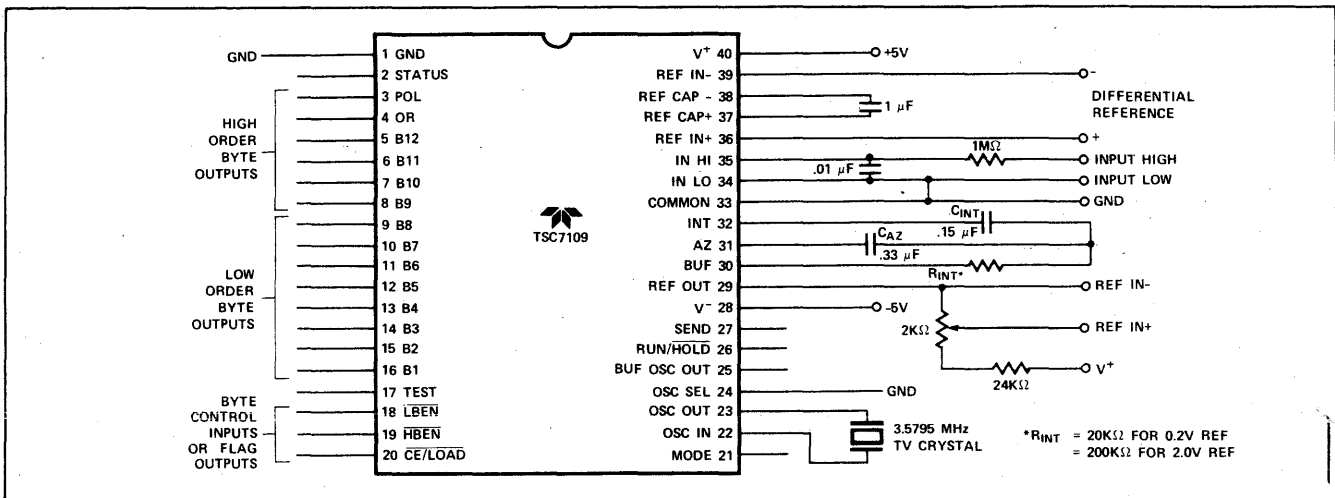
Part No.	Package	Temp. Range
TSC7109CPL	40-Pin Plastic Dip	0°C to +70°C
TSC7109BCPL	40-Pin Plastic Dip	0°C to +70°C
TSC7109IJL	40-Pin CerDIP	-25°C to +85°C
TSC7109BIJL	40-Pin CerDIP	-25°C to +85°C
TSC7109MJL	40-Pin CerDIP	-55°C to +125°C

### Features

- 12-Bit Plus Sign Integrating A/D Converter with Overrange Indication
- Sign Magnitude Coding Format
- True Differential Signal Input and Differential Reference Input
- Low Noise — Typically 15  $\mu\text{V}$ p-p
- High Normal Mode Noise and Line Frequency Rejection
- 1 pA Typical Input Current
- No Zero Adjustment
- TTL Compatible Byte Organized Tri-State Outputs
- UART Handshake Mode for Simple Serial Data Transmission
- Direct Bus Connection for 8 or 16-Bit Bus — 3.58 MHz Crystal Provides 7.5 Conversions Per Second for 60 Hz Rejection — External RC Network Provides up to 30 Conversions Per Second
- Power Dissipation Typically Less Than 20 mW
- Internal Voltage Reference

Part No.	Package	Temp. Range
TSC7109CBQ	60-Pin Plastic Flat Package: Formed Leads	0°C to +70°C
TSC7109CSQ	60-Pin Plastic Flat Package: Unformed Leads	0°C to +70°C
<b>Devices Available with 160 Hour, +125°C Burn-In</b>		
TSC7109CPL/BI	40-Pin Plastic Dip	0°C to +70°C
TSC7109IJL/BI	40-Pin CerDIP	-25°C to +85°C
<b>Devices with MIL-STD-883 Processing</b>		
TSC7109MJL/883	40-Pin CerDIP	-55°C to +125°C

### Test Circuit (See Figure 1 for typical connection to A UART or Microcomputer)



## General Description

The TSC7135 4 1/2 digit analog converter offers 50 ppm (1 part in 20,000) resolution with a maximum linearity error of 1 count. An auto-zero cycle reduces the zero error to below 10  $\mu$ V and zero drift to 0.5  $\mu$ V/ $^{\circ}$ C. Source impedance error sources are minimized by a 10 pA maximum input current. Rollover error is limited to  $\pm 1$  count.

By combining the TSC7135 with a TSC7211A (LCD), TSC7212A (LED) or TSC700A (High LED Segment Current) driver a 4 1/2 digit display DVM or DPM can be constructed. Overrange and underrange signals support automatic range switching and special display blanking/flashing applications.

Micro-processor based measurement systems are supported by the TSC7135 Busy, Strobe and Run/HOLD control signals. Remote data acquisition systems with data transfer via UARTs are also possible. The additional control pins and multiplexed BCD outputs make the TSC7135 the ideal converter for display or  $\mu$ -processor based measurement systems.

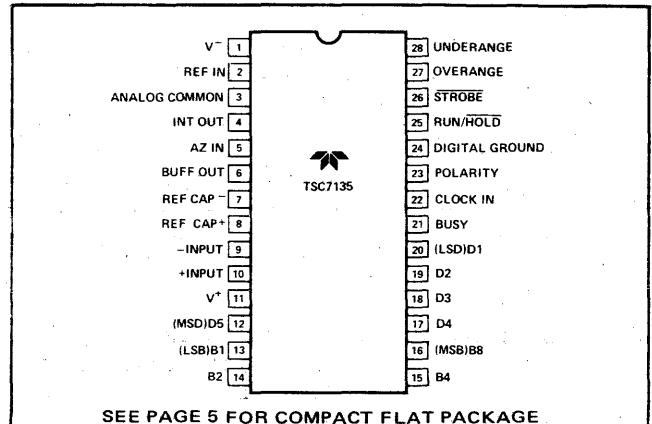
## Features

- Low Rollover Error .....  $\pm 1$  Count Maximum
- Guaranteed  $\pm 1$  Count Maximum Error
- Guaranteed Zero Reading for 0 V Input
- True Polarity Indication at Zero for Null Detection
- Multiplexed BCD Data Output
- TTL Compatible Outputs
- Differential Input
- Control Signals Permit Interface to UARTS and  $\mu$ -Processors
- Auto-ranging Supported with Over and Underrange Signals
- Blinking Display Visually Indicates Overrange Condition
- Low Input Current ..... 1 pA
- Low Zero Reading Drift ..... 2  $\mu$ V/ $^{\circ}$ C
- Interface to TSC7211A, TSC7212A, and TSC700A Display Drivers
- Available in Compact Flat Package

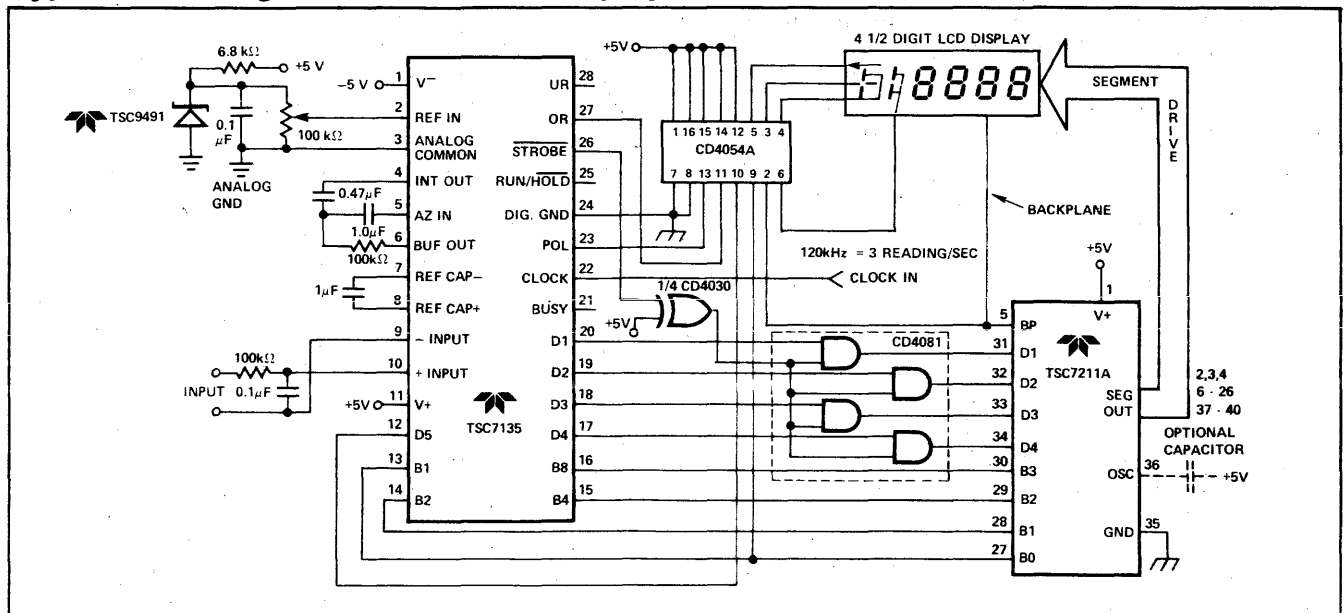
## Ordering Information

Part No.	Package	Temperature Range
TSC7135CJI	28-Pin CerDIP	0 $^{\circ}$ C to +70 $^{\circ}$ C
TSC7135CPI	28-Pin Plastic	0 $^{\circ}$ C to +70 $^{\circ}$ C
TSC7135CBQ	60-Pin Plastic Flat Package w/ Formed Leads	0 $^{\circ}$ C to +70 $^{\circ}$ C
TSC7135CSQ	60-Pin Plastic Flat Package w/ Unformed Leads	0 $^{\circ}$ C to +70 $^{\circ}$ C

## Pin Configuration



## Typical 4 1/2 Digit DVM with LCD Display



Preliminary

## TSC7650 Precision Chopper-Stabilized Operational Amplifier

- 0.7  $\mu\text{V}$  Offset Voltage
- 50  $\text{nV}/^\circ\text{C}$  Offset Voltage Drift

### General Description

The TSC7650 CMOS chopper-stabilized operational amplifier practically removes offset voltage error terms from system error calculations. The 5  $\mu\text{V}$  maximum  $V_{OS}$  specification, for example, represents a fifteen times improvement over the industry standard OPO7E. The 50  $\text{nV}/^\circ\text{C}$  offset drift specification is over twenty-five times lower than the OPO7E. The increased performance eliminates  $V_{OS}$  trim procedures, periodic potentiometer adjustment and the reliability problems caused by damaged trimmers.

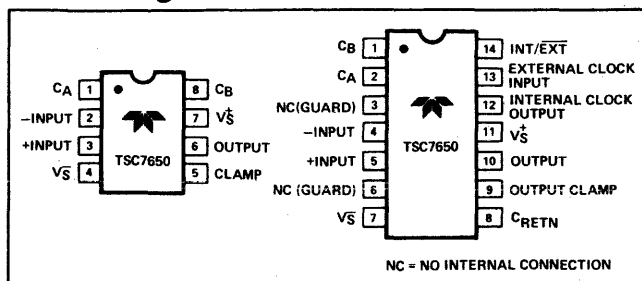
The TSC7650 performance advantages are achieved without the additional manufacturing complexity and cost incurred with laser or "zener zap"  $V_{OS}$  trim techniques. The TSC7650 is one of the lowest cost precision operational amplifiers available.

The TSC7650 nulling scheme corrects both dc  $V_{OS}$  errors and  $V_{OS}$  drift errors with temperature. A nulling amplifier alternately corrects its own  $V_{OS}$  errors and the main amplifier  $V_{OS}$  error. Offset nulling voltages are stored on two user supplied external capacitors. The capacitors connect to the internal amplifier  $V_{OS}$  null points. The main amplifier input signal is never switched. Switching spikes are not present at the TSC7650 output. The null scheme keeps  $V_{OS}$  errors low throughout the operating temperature range. Laser and "zener zap" trimming can correct for  $V_{OS}$  at only one temperature.

The nulling circuit oscillator and control circuits are integrated on chip. Only two external  $V_{OS}$  error storage capacitors are required. The TSC7650 operates as a conventional operational amplifier with vastly improved input specifications. The low  $V_{OS}$  and  $V_{OS}$  drift errors make the TSC7650 ideal for thermocouple, thermistor, and strain gauge applications. Low dc errors and high open loop gain make the TSC7650 an excellent preamplifier for precision analog to digital converters like the TSC7135 and TSC800.

The 14-pin dual in line package (DIP) has an external oscillator input to drive the nulling circuitry for optimum noise performance. Both the 8 and 14-pin DIP have an output voltage clamp circuit to minimize overload recovery time.

### Pin Configuration



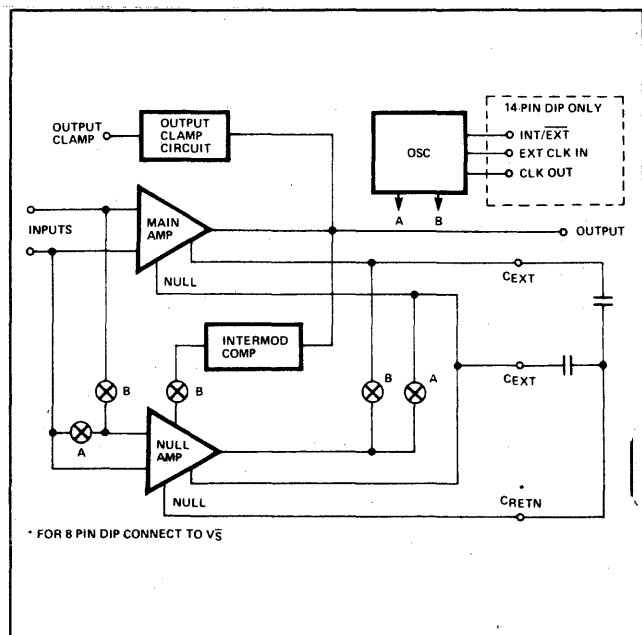
### Features

- Low Input Offset Voltage ..... 0.7  $\mu\text{V}$
- Low Input Offset Voltage Drift ..... 0.05  $\mu\text{V}/^\circ\text{C}$  Max.
- Low Input Bias Current ..... 10 pA Max.
- High Impedance Differential CMOS Inputs .....  $10^{12} \Omega$
- High Open Loop Voltage Gain ..... 120 dB Min.
- Low Input Noise Voltage ..... 2.0  $\mu\text{Vp-p}$
- High Slew Rate ..... 2.5  $\text{V}/\mu\text{s}$
- Low Power Operation ..... 20 mW
- Output Clamp Speeds Recovery Time
- Compensated Internally for Stable Unity Gain Operation
- Direct Replacement for ICL7650
- Available in 8-Pin Dip

### Ordering Information

Part No.	Package	Temp. Range	Max. Offset Voltage
TSC7650CPA	8-Pin Plastic Dip	0°C to +70°C	5 $\mu\text{V}$
TSC7650IJA	8-Pin CerDIP	-25°C to +85°C	5 $\mu\text{V}$
TSC7650CPD	14-Pin Plastic Dip	0°C to +70°C	5 $\mu\text{V}$
TSC7650IJD	14-Pin CerDIP	-25°C to +85°C	5 $\mu\text{V}$

### Functional Diagram





### General Description

The TSC7660 DC to DC converter will generate a negative voltage from a positive source. With two external capacitors the TSC7660 will convert a 1.5 V to 10.0 V input signal to -1.5 V to -10.0 V level. The TSC7660 easily generates -5 V in +5 V digital systems.

Many analog to digital converters, digital to analog converters, operational amplifiers, and multiplexers require negative supply voltages. The TSC7660 allows +5 V digital logic systems to incorporate these analog components without adding an additional main power source. The TSC7660 can lower total system cost, ease engineering development and save space, power and weight.

The TSC7660 charges a capacitor to the applied supply voltage. Internal analog gates connect the capacitor across the output. Charge is transferred to an output storage capacitor completing the voltage conversion. Operation requires only two external capacitors for supply voltage <6.5 V.

Contained on-chip are a series DC power supply regulator, RC oscillator, voltage level translator, four output power MOS switches, and a unique logic element which senses the most negative voltage in the device and ensures that the output N-channel switches are not forward biased. This assures latch-up free operation.

The oscillator, when unloaded, oscillates at a nominal frequency of 10 kHz for an input supply voltage of 5.0 volts. This frequency can be lowered by the addition of an external capacitor to the "OSC" terminal, or the oscillator may be overdriven by an external clock.

The "LV" terminal may be tied to GROUND to bypass the internal series regulator and improve low voltage (LV) operation. At medium to high voltages (+3.5 to +10.0 volts), the LV pin is left floating to prevent device latchup.

### Features

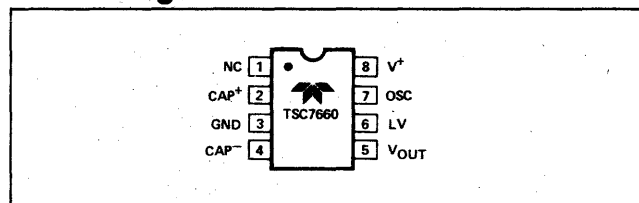
- Converts +5 V Logic Supply to  $\pm 5$  V System
- Wide Input Voltage Range ..... 1.5 V to 10.0 V
- Efficient Voltage Conversion ..... 99.9%
- Excellent Power Efficiency ..... 98%
- Low Supply Current ..... 500  $\mu$ A Max.
- Cascade for Output Voltage Multiplication
- Low Cost and Easy to Use
  - Only 2 External Capacitors Required

### Ordering Information

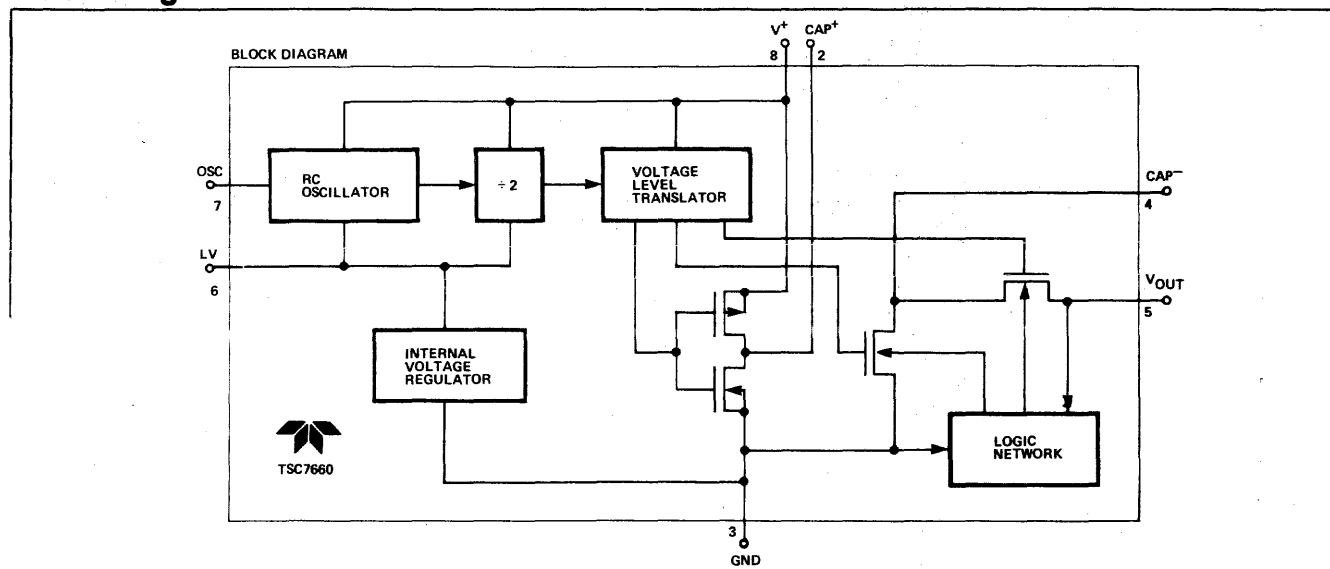
Part No.	Package	Temperature Range
TSC7660CPA	8-Pin Plastic Dip	0°C to +70°C
TSC7660IJA	8-Pin CerDIP	-40°C to +85°C
TSC7660MJA	8-Pin CerDIP	-55°C to +125°C
TSC7660/Y	Chip	25°C
Devices with MIL-STD-883 Processing		
TSC7660MJA/883	8-Pin CerDIP	-55°C to +125°C

The TSC7660 open circuit output voltage is equal to the input voltage to within 0.1%. The TSC7660 has a 98% power conversion efficiency for a 2 -5 mA load currents.

### Pin Configuration



### Block Diagram



### General Description

The TSC800 is a 15-bit plus sign integrating analog to digital converter. The TSC800 improves the conventional two cycle dual slope conversion cycle by incorporating system zero and integrator output zero phases. Offset error sources are automatically zeroed and overrange recovery time is reduced. The integrating conversion technique is immune to the noise spikes that introduce conversion errors in successive approximation converters.

The externally adjustable clock allows integration periods which are integral multiples of 50 Hz or 60 Hz for maximum power-line noise rejection. By using the 2.4576 MHz crystal oscillator mode (2.5 CONV/SEC) 50, 60 and 400 Hz signals are rejected.

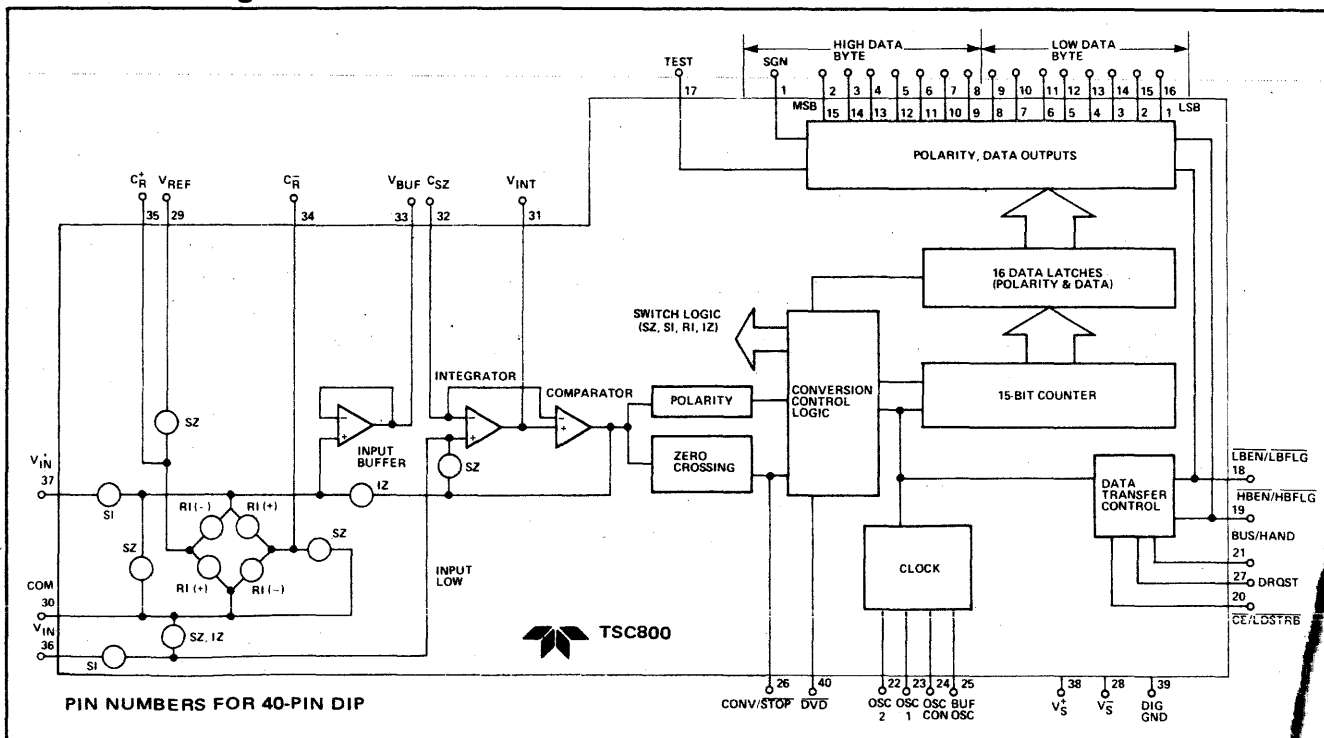
Micro-processor interface signals support single byte (16-bit) or two byte (8-bit) parallel data transfers. A "handshake" operating mode supports serial data transmission via a UART. A serial count output is derivable by gating the clock signal with data valid (DVD). The count output pulses may be used in serial fiber optic transmission systems.

The high impedance differential inputs, 5 pA input leakage current, 16-bit dynamic range and interface control signals make the high resolution TSC800 the ideal analog to digital converter for process control, data logging and "intelligent" measurement systems.

### Features

- 15 Bit Resolution. Plus Sign Bit
  - 96 dB Dynamic Range
- Integrating Dual Slope Converter
  - Monotonic
  - Eliminate 50/60 Hz "Line" Interference
  - High Noise Immunity
  - Auto Zero Cycle Eliminates Trimming
  - Incorporates Integrator Zero Cycle for Fast Overload Recovery
- Three State Data Bit/Sign Outputs
  - 8 or 16 Bit Parallel Data Transfer to  $\mu$ -Processor Bus
- UART Control Signals
  - Serial Data Transmission
  - "Handshake" Data Transfer
  - Distributed Control Systems
  - Fiber Optic Transmission Systems
- Easy Conversion Cycle Monitoring and Control
  - Data Valid Output Signal
  - Continuous or Convert on Command Operation
- High Impedance Differential Input
  - 15 pA Maximum Input Current
- Low Input Noise
  - $15 \mu V_{p-p}$
- On Chip Crystal Oscillator for 2.5 Conversions/Sec.
  - $f_{xtal} = 2.4576 \text{ MHz}$
  - 100 mSEC Integration Period Rejects 50, 60, 400 Hz Interference Signals
- Convenient  $\pm 5 \text{ V}$  Supply Operation
  - Low Power Dissipation ..... 20 mW
- Static Discharge Protected Inputs
- Available in 60-Pin Flat Package

### Functional Diagram



*Advance  
Product  
Information*

INTERFACE

Teledyne Semiconductor

**General Description**

The TSC805 is a 3 1/2 digit integrating analog-to-digital converter with tri-plex LCD display drive and automatic ranging. The CMOS TSC805 contains all the logic and analog switches needed to manufacture an auto-ranging instrument for ohms and voltage measurements.

The auto-ranging feature automatically selects the proper display decimal point location. Auto-ranging is available during ohms (high and low voltage) and voltage (AC & DC) measurements. Auto-ranging eliminates expensive range switches in hand-held DMM designs and makes compact meters easier and less costly to design. Auto-ranging instruments are easy to operate and are ideal for non-technical end users. The auto-range feature may be bypassed allowing decimal point selection and input attenuator selection control through a single line input.

The TSC805 includes an AC to DC converter for AC audio frequency range measurements. Only inexpensive external diodes/resistors/capacitors are required.

A complete LCD annunciator set visually describes the TSC805 meter function and measurement range during ohms, voltage and current operation. AC or DC measurements are indicated. A low battery detection circuit also sets a low battery warning display annunciator.

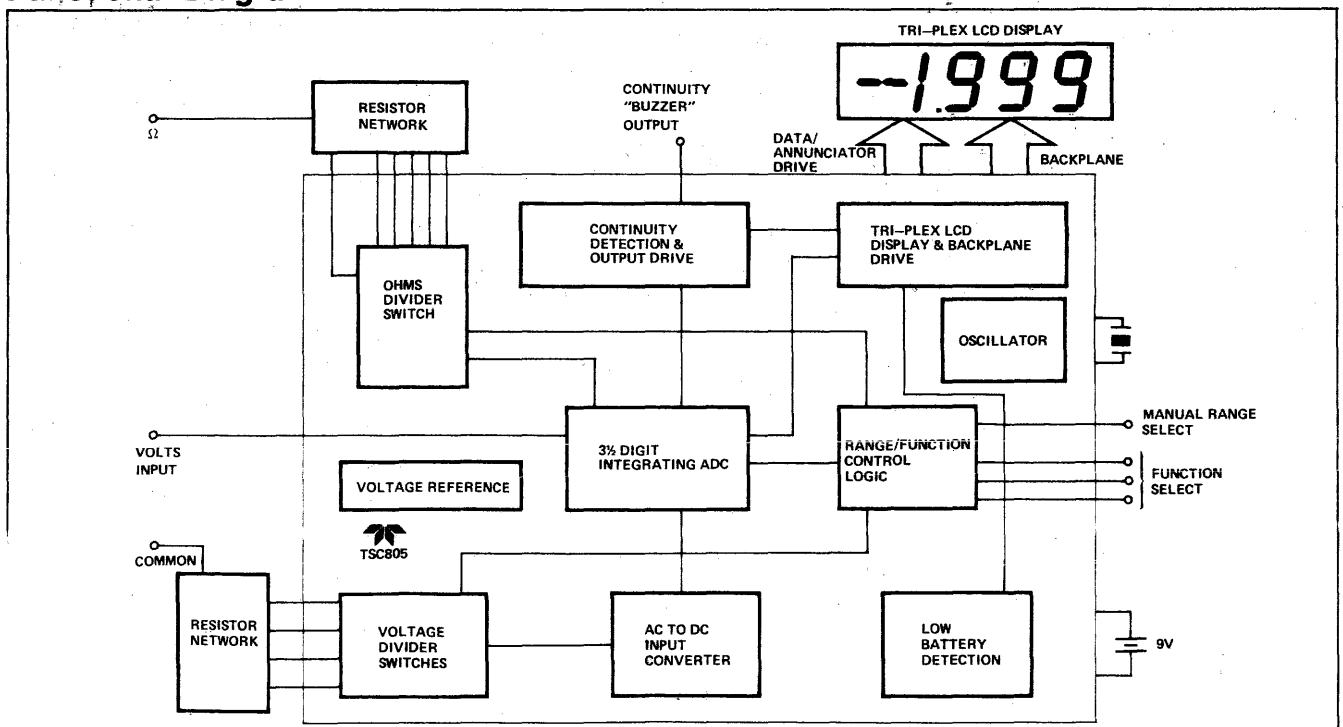
**Features**

- Auto-Ranging on volts (AC & DC) and Ohms Scale
- Automatic Sign, Function and Unit Display
- AC to DC Voltage Converter on Chip
- Low Voltage Ohms Range Option
- Continuity Detection & "Buzzer" Drive Output
- 3 1/2 Digit Resolution on 200 mV Full-Scale Range
- Low Noise Input Stage ..... 15  $\mu$ Vp-p
- Precision Internal Reference ..... 50 ppm/ $^{\circ}$ C Max.
- Triplex LCD Drivers On Chip
- Low Battery Detection & LCD Annunciator
- Convenient 9 V Battery Operation
- Low Power Consumption ..... 10 mW
- Compact 60-Pin Flat Package

The "low ohms" measurement option allows in circuit resistance measurements by preventing semiconductor junctions from being forward biased. A zero ohms adjust is included to compensate for lead resistance.

A continuity buzzer output is activated with inputs less than 5% of full-scale. An overrange input also enables the buzzer and flashes the MSD display. Offering single 9 V battery operation, 10 mW power consumption, a precision internal voltage reference (50 ppm/ $^{\circ}$ C max. TC) and available in a compact 60-pin flat package the TSC805 is ideal for portable instruments.

**Functional Diagram**



### General Description

The TSC826 is a CMOS analog-to-digital converter that directly drives liquid crystal bar graph displays. LCD drivers are on-chip for forty data segment, zero, polarity, and over-range annunciators. A backplane oscillator and driver are included. All active components are on-chip for a 2.5% resolution bar graph display. Up to 16 conversions per second are possible.

An auto-zero cycle guarantees a zero display for zero volt input. No adjustment potentiometers are needed. The precision internal reference with a 35 ppm/°C temperature coefficient virtually eliminates full scale errors over temperature.

The CMOS TSC826 draws less than 100 μA from a 9 V battery guaranteeing long life in portable applications. Available in a 60-pin flat package compact portable designs with multiple bar graph indicators are possible.

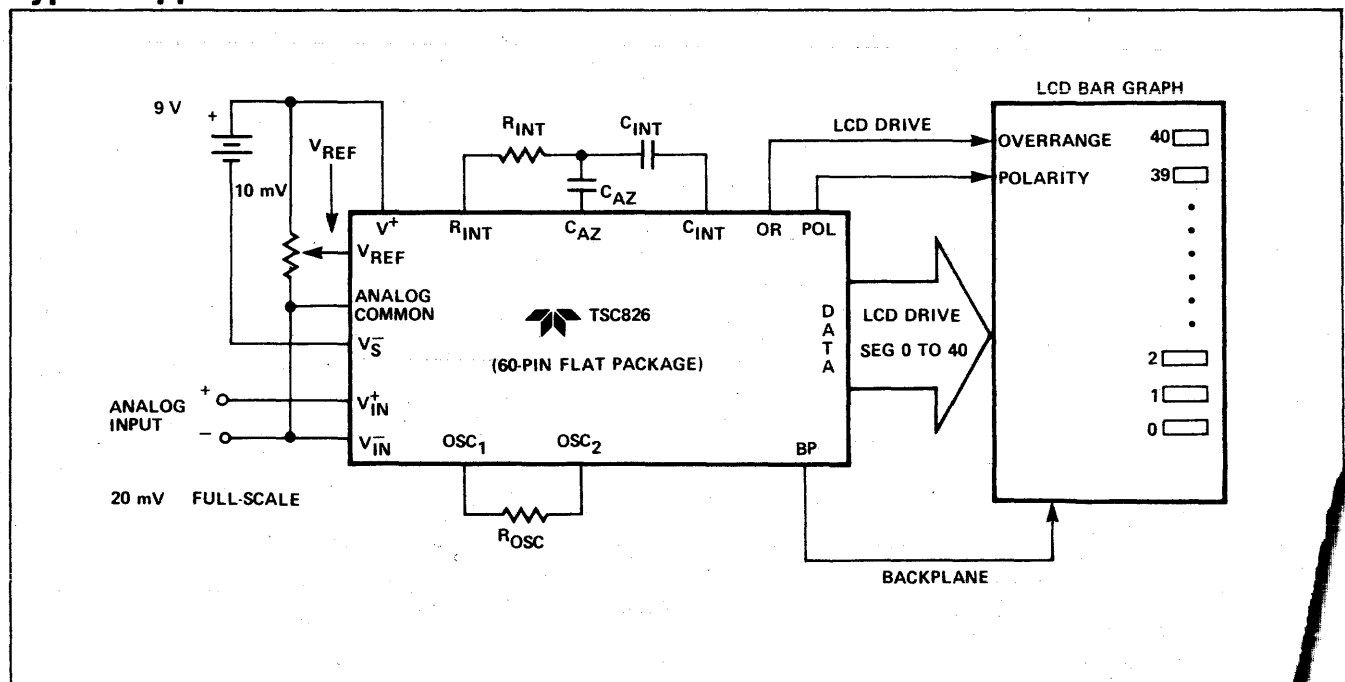
### Features

- Bar Graph Data Display . . . 40 Segments plus zero
- Overrange Plus Polarity Indication
- Precision On-Chip Reference . . . 35 ppm/°C
- Auto-Zero Cycle Eliminates Zero Adjust Potentiometer
- Low Power Consumption . . . 1 mW
- Differential Analog Input
- LCD Drivers and Backplane Oscillator
- 9 V Battery Operation
- 20 mV Full-Scale Operation
- Compact Flat Package

### Preliminary Ordering Information

Part No.	Package	Temperature Range
TSC826IBQ	60-Pin Plastic Flat Package (Formed Leads)	-40°C to +85°C

### Typical Application



## Serial Input/16-Bit Parallel Output Peripheral Driver

- High Voltage, High Current Outputs

### General Description

The Teledyne Semiconductor TSC9403 and TSC9404 are serial input, 16-bit parallel output shift registers. High output power MOS switching transistors make the TSC9403 and TSC9404 ideal interface circuits between microprocessor I/O ports and high current/voltage peripherals. The CMOS construction limits quiescent power dissipation to 20 mW.

The TSC9403 common source, open drain MOS outputs sustain 20 V in the OFF state and maintain leakage currents under 100  $\mu$ A. The TSC9404 outputs are rated at 15 V. The 16 parallel outputs will continuously sink 60 mA. ( $V_{SAT} \leq 0.5$  V).

Successive connection of serial data outputs to serial data inputs make longer length serial to parallel conversions possible. Device cascading makes the TSC9403 and TSC9404 ideal thermal printhead or high resolution LED bar graph drivers.

### Features

- High Voltage Outputs: 20 V (TSC9403), 15 V (TSC9404)
- High Output Current Sink Capability: 60 mA
- Low Standby Power: 20 mW
- High Speed Operation: 3.0 MHz
- 16 Parallel Outputs
- Cascading Possible for Longer Data Words

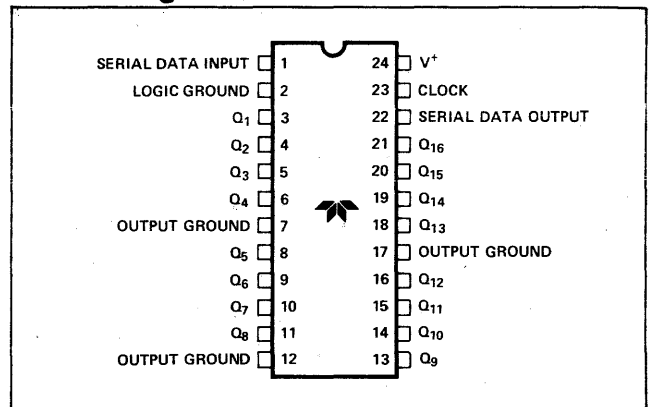
### Applications

- Incandescent Lamp Driver
- Thermal Printhead Driver
- LED Bar Graph Driver
- High Current, Microprocessor Serial Port Expander
- Relay/Solenoid Driver
- Tungsten Lamp Driver
- SCR Gate Driver

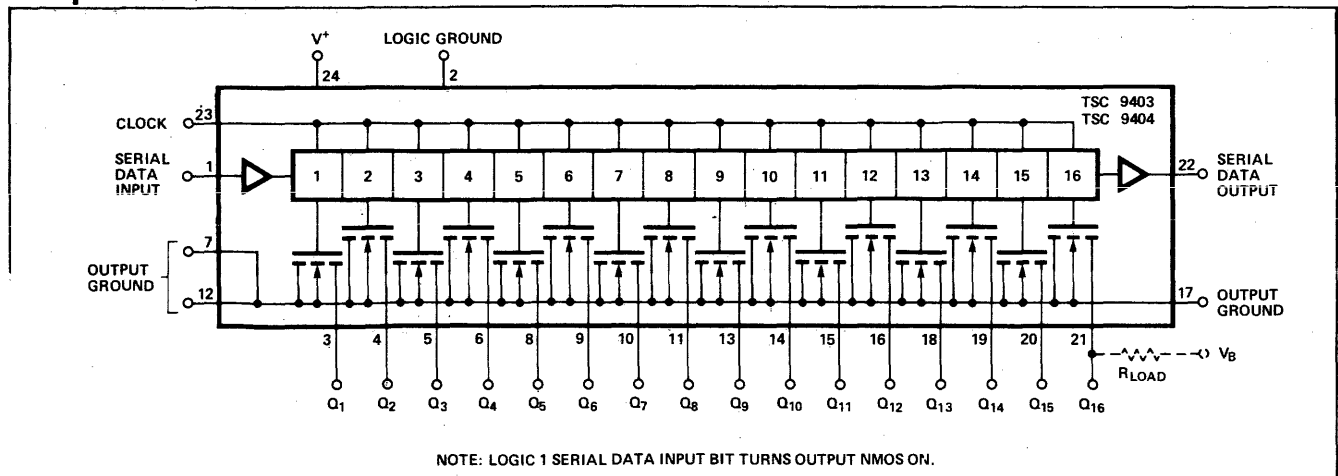
### Ordering Information

Part	Package	Temperature Range	Output Voltage
TSC9403CJ	24-Pin Epoxy Dip	0°C to 70°C	20 V
TSC9403IL	24-Pin CerDIP	-25°C to 85°C	20 V
TSC9404CJ	24-Pin Epoxy Dip	0°C to 70°C	15 V
TSC9404IL	24-Pin CerDIP	-25°C to 85°C	15 V
TSC9404ML	24-Pin CerDIP	-55°C to 125°C	15 V
<b>Devices Available with MIL-STD-883 Processing</b>			
TSC9404ML/883	24-Pin CerDIP	-55°C to 125°C	15 V

### Pin Configuration



### Simplified Schematic





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- LinCMOSTM Technology
- 8-Bit Resolution A/D Converter
- On-Chip 12-Channel Analog Multiplexer
- Built-In Self-Test Mode
- Software-Controllable Sample and Hold
- Total Unadjusted Error . . .  $\pm 0.5$  LSB Max
- Direct Replacement for Motorola MC145040

TYPICAL PERFORMANCE:	TLC540	TLC541
Channel Acquisition Time	2 $\mu$ s	7 $\mu$ s
Conversion Time	10 $\mu$ s	19 $\mu$ s
Sampling Rate	$71 \times 10^3$	$29 \times 10^3$
Power Dissipation	6 mW	6 mW

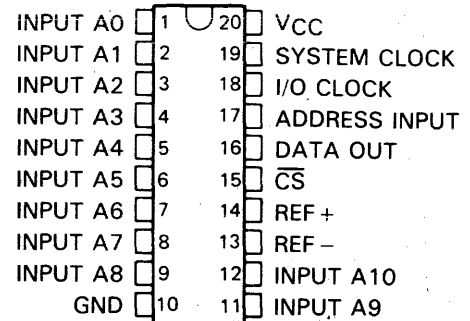
**description**

The TLC540 and TLC541 are LinCMOSTM A/D peripherals built around an 8-bit switched-capacitor successive-approximation A/D converter. They are designed for serial interface to a microprocessor or peripheral via a three-state output with up to four control inputs [including independent System Clock, I/O Clock, Chip Select ( $\overline{CS}$ ), and Address Input]. A 4-megahertz system clock for the TLC540 and a 2.1-megahertz system clock for the TLC541 with a design that includes simultaneous read/write operation allow high-speed data transfers and sample rates of up to 71,910 samples per second for the TLC540 and 29,144 samples per second for the TLC541. In addition to the high-speed converter and versatile control logic, there is an on-chip 12-channel analog multiplexer that can be used to sample any one of 11 inputs or an internal "self-test" voltage, and a sample-and-hold that can operate automatically or under processor control.

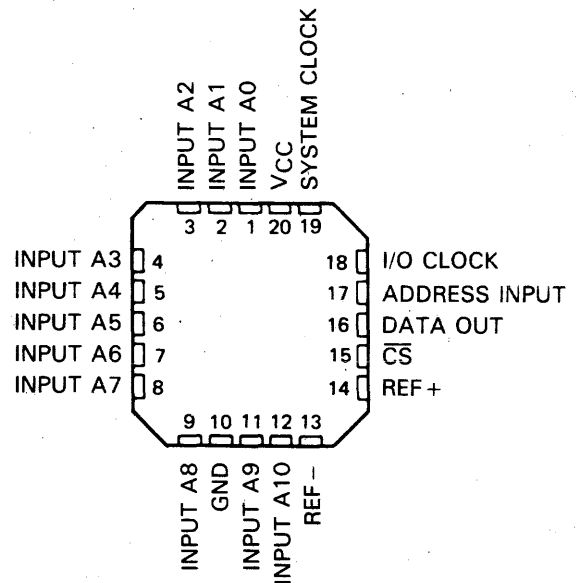
The converters incorporated in the TLC540 and TLC541 feature differential high-impedance reference inputs that facilitate ratiometric conversion, scaling, and analog circuitry isolation from logic and supply noises. A totally switched-capacitor design allows guaranteed low-error ( $\pm 0.5$  LSB) conversion in 10 microseconds for the TLC540 and 19 microseconds for the TLC541 over the full operating temperature range.

The TLC540M and the TLC541M are characterized for operation over the full military temperature range of  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ . The TLC540I and the TLC541I are characterized for operation from  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

J OR N DUAL-IN-LINE PACKAGE  
(TOP VIEW)



FK OR FN PACKAGE  
(TOP VIEW)



**ADVANCE INFORMATION**

This document contains information on a new product. Specifications are subject to change without notice.

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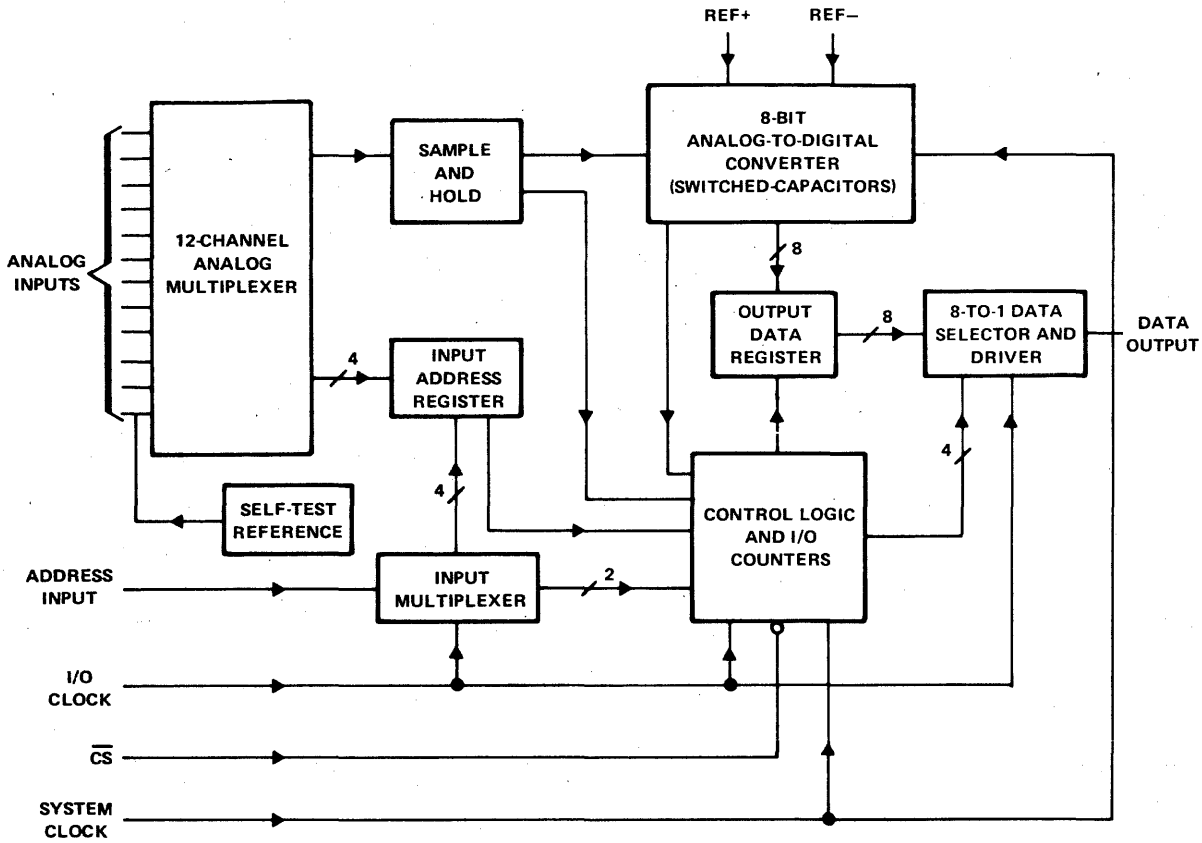


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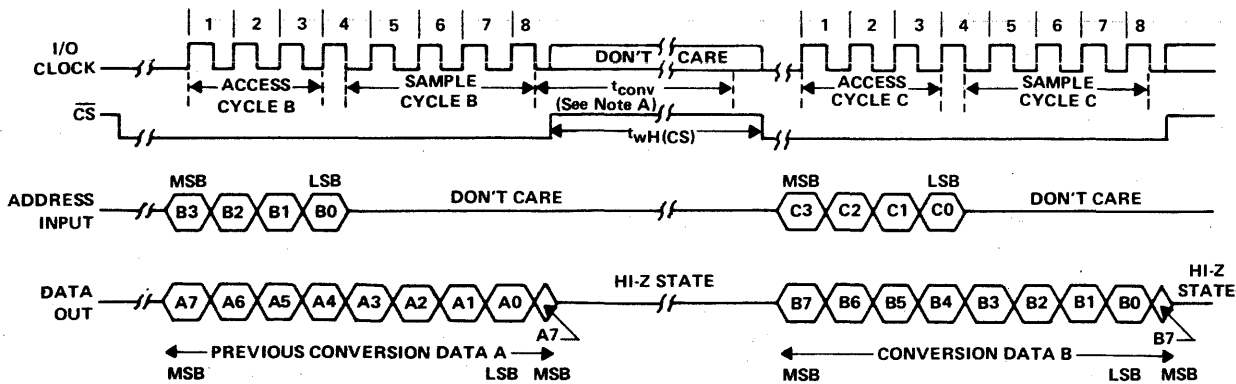
# TYPES TLC540M, TLC540I, TLC541M, TLC541I

## 8-BIT ANALOG-TO-DIGITAL PERIPHERALS WITH SERIAL CONTROL AND 11 INPUTS

functional block diagram



operating sequence



NOTE A: The conversion cycle, which requires 40 system clock periods, is initiated with the 8th I/O clock  $\downarrow$  after CS  $\downarrow$  for the channel whose address exists in memory at that time.

TEXAS  
INSTRUMENTS

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# TYPES TLC540M, TLC540I, TLC541M, TLC541I 8-BIT ANALOG-TO-DIGITAL PERIPHERALS WITH SERIAL CONTROL AND 11 INPUTS

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1)	6.5 V
Input voltage range (any input)	-0.3 V to $V_{CC} + 0.3$ V
Output voltage range	-0.3 V to $V_{CC} + 0.3$ V
Peak input current range (any input)	$\pm 10$ mA
Peak total input current (all inputs)	$\pm 30$ mA
Continuous total dissipation at 25°C (see Note 2)	875 mW
Operating free-air temperature range: TLC540I, TLC541I	-40°C to 85°C
TLC540M, TLC541M	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: FK, FN, or N package	260°C

- NOTES: 1. All voltage values are with respect to digital ground with REF- and GND wired together (unless otherwise noted).  
 2. For operation above 25°C free-air temperature, see Dissipation Derating Curves, Section 2. In the J package, TLC540M and TLC541M chips are alloy mounted, TLC540I and TLC541I chips are glass mounted.

**recommended operating conditions**

	TLC540			TLC541			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.75	5	5.5	4.75	5	5.5	V
Positive reference voltage, $V_{REF+}$ (see Note 3)	1.25	$V_{CC}$	$V_{CC}+0.1$	1.25	$V_{CC}$	$V_{CC}+0.1$	V
Negative reference voltage, $V_{REF-}$ (see Note 3)	-0.1	0	$V_{CC}-1.25$	0.1	0	$V_{CC}-1.25$	V
Differential reference voltage, $V_{REF+} - V_{REF-}$ (see Note 3)	1	$V_{CC}$	$V_{CC}+0.2$	1	$V_{CC}$	$V_{CC}+0.2$	V
Analog input voltage (see Note 3)	0		$V_{CC}$	0		$V_{CC}$	V
High-level control input voltage, $V_{IH}$	2			2			V
Low-level control input voltage, $V_{IL}$			0.8			0.8	V
Setup time, address bits at data input before I/O CLK $\uparrow$ , $t_{su(A)}$	200			400			ns
Setup time, $\overline{CS}$ low before clocking in first address bit, $t_{su(CS)}$ (see Note 4)	2			2			System clock cycles
Input/Output clock frequency, $f_{CLK(I/O)}$	0.005		2.048	0		0.525	MHz
System clock frequency, $f_{CLK(SYS)}$				$f_{CLK(I/O)}$			MHz
System clock high, $t_{wH(SYS)}$	110			210			ns
System clock low, $t_{wL(SYS)}$	100			190			ns
Input/Output clock high, $t_{wH(I/O)}$	200			808			ns
Input/Output clock low, $t_{wL(I/O)}$	200			808			ns
Clock transition time (see Note 5)	System	$f_{CLK(SYS)} \leq 1048$ kHz	30		30		ns
		$f_{CLK(SYS)} > 1048$ kHz	20		20		
	I/O	$f_{CLK(I/O)} \leq 525$ kHz	100		100		ns
		$f_{CLK(I/O)} > 525$ kHz	40		40		
Operating free-air temperature, $T_A$	TLC540M, TLC541M	-55	125	-55	125	°C	
	TLC540I, TLC541I	-40	85	-40	85		

- NOTES: 3. Analog input voltages greater than that applied to REF+ convert as all '1's (11111111), while input voltages less than that applied to REF- convert as all '0's (00000000). For proper operation, REF+ voltage must be at least 1 volt higher than REF- voltage. Also, adjusted errors may increase as this differential reference voltage falls below 4.75 volts.  
 4. To minimize errors caused by noise at the Chip Select input, the internal circuitry waits for two system clock cycles (or less) after a chip select falling edge is detected before responding to control input signals. Therefore, no attempt should be made to clock-in address data until the chip select setup time has elapsed.  
 5. This is the time required for the clock input signal to fall from  $V_{IH}$  min to  $V_{IL}$  max or to rise from  $V_{IL}$  max to  $V_{IH}$  min.

# TYPES TLC540M, TLC540I, TLC541M, TLC541I

## 8-BIT ANALOG-TO-DIGITAL PERIPHERALS WITH SERIAL CONTROL AND 11 INPUTS

electrical characteristics over recommended operating temperature range,  
 $V_{CC} = V_{REF+} = 4.75\text{ V to }5.5\text{ V}$  (unless otherwise noted),  $f_{CLK(I/O)} = 2.028\text{ MHz}$  for TLC540 or  $f_{CLK(I/O)} = 0.525\text{ MHz}$  for TLC541

PARAMETER		TEST CONDITIONS	MIN	TYP <sup>†</sup>	MAX	UNIT
$V_{OH}$	High-level output voltage (pin 16)	$V_{CC} = 4.75\text{ V}$ , $I_{OH} = 360\text{ }\mu\text{A}$	2.4			V
$V_{OL}$	Low-level output voltage	$V_{CC} = 4.75\text{ V}$ , $I_O = 3.2\text{ mA}$			0.4	V
$I_{OZ}$	Off-state (high-impedance state) output current	$V_O = V_{CC}$ , $\overline{CS}$ at $V_{CC}$			10	$\mu\text{A}$
		$V_O = 0$ , $\overline{CS}$ at $V_{CC}$			-10	
$I_{IH}$	High-level input current	$V_I = V_{CC} + 0.3\text{ V}$		0.005	2.5	$\mu\text{A}$
$I_{IL}$	Low-level input current	$V_I = 0$		-0.005	-2.5	$\mu\text{A}$
$I_{CC}$	Operating supply current	$\overline{CS}$ at 0 V		1.2	2	mA
	Selected channel leakage current	Selected channel at $V_{CC}$ , Unselected channel at 0 V		0.4	1	$\mu\text{A}$
		Selected channel at 0 V, Unselected channel at $V_{CC}$		-0.4	-1	
$I_{CC} + I_{REF}$	Supply and reference current	$V_{REF+} = V_{CC}$ , $\overline{CS}$ at 0 V		1.3	3	mA
$C_i$	Input capacitance	Analog inputs		7	55	pF
		Control inputs		5	15	

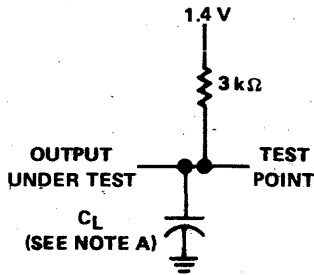
<sup>†</sup>All typical values are at  $T_A = 25^\circ\text{C}$ .

operating characteristics over recommended operating free-air temperature range,  
 $V_{CC} = V_{REF+} = 4.75\text{ V to }5.5\text{ V}$ ,  $f_{CLK(I/O)} = 2.048\text{ MHz}$  for TLC540 or  $0.525\text{ MHz}$  for TLC541,  $f_{CLK(SYS)} = 4\text{ MHz}$  for TLC540 or  $2.097\text{ MHz}$  for TLC541.

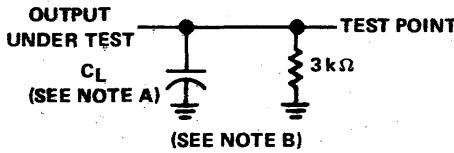
PARAMETER	TEST CONDITIONS	TLC540			TLC541			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
Linearity error	See Note 6			$\pm 0.5$			$\pm 0.5$	LSB
Zero error	See Note 7			$\pm 0.5$			$\pm 0.5$	LSB
Full-scale error	See Note 7			$\pm 0.5$			$\pm 0.5$	LSB
Total unadjusted error	See Note 8			$\pm 0.5$			$\pm 0.5$	LSB
Self-test output code	input address = 1011 (A11) (See Note 9)	01111101 (125)		10000011 (131)	01111101 (125)		10000011 (131)	
$t_{conv}$	Conversion time			10			19	$\mu\text{s}$
$t_{acq}$	Channel acquisition time			4			4	I/O clock cycles
$t_v$	Time output data remains valid after I/O clock $\downarrow$	10			10			ns
$t_d$	Delay time, I/O clock $\downarrow$ to data output valid			200			400	ns
$t_{acc}$	Output access time (delay to valid output after chip select $\downarrow$ )	1		3	1		3	System clock cycles
$t_{en}$	Output enable time			150			150	ns
$t_{dis}$	Output disable time			150			150	ns
$t_{r(bus)}$	Data bus rise time			300			300	ns
$t_{f(bus)}$	Data bus fall time			300			300	ns

- NOTES: 6. Linearity error is the maximum deviation from the best straight line through the A/D transfer characteristics.  
 7. Zero Error is the difference between the output of an ideal and an actual A/D for zero input voltage; full-scale error is that same difference for full-scale input voltage.  
 8. Total Unadjusted Error is the sum of linearity, zero, and full-scale errors.  
 9. Both the input address and the output codes are expressed in positive logic.

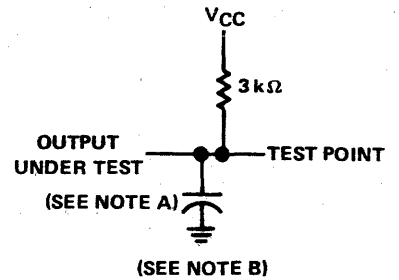
PARAMETER MEASUREMENT INFORMATION



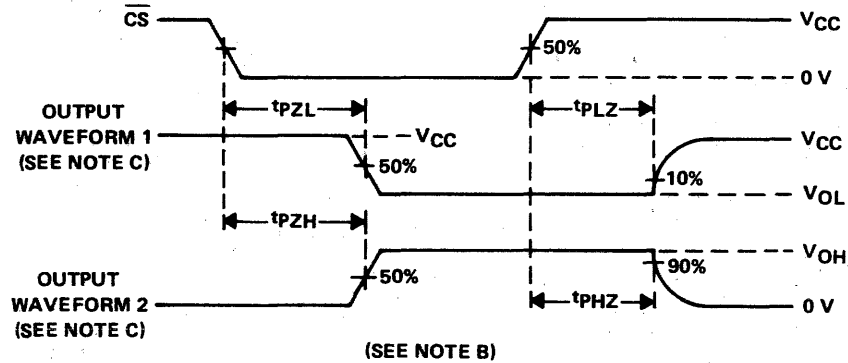
LOAD CIRCUIT FOR  
 $t_d$ ,  $t_{acc}$ ,  $t_r$ ,  $t_f$



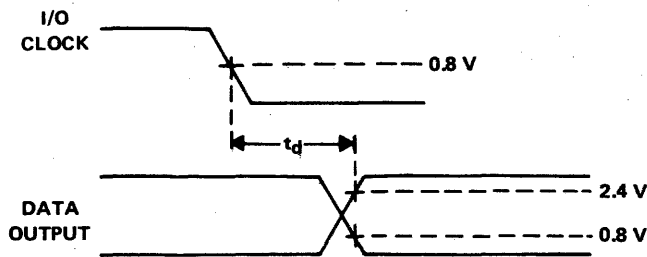
LOAD CIRCUIT FOR  
 $t_{pZH}$  AND  $t_{pHZ}$



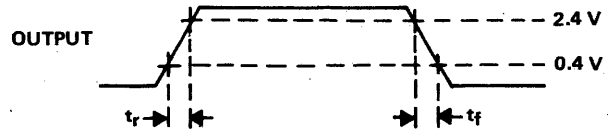
LOAD CIRCUIT FOR  
 $t_{pZL}$  AND  $t_{pLZ}$



VOLTAGE WAVEFORMS FOR ENABLE AND DISABLE TIMES



VOLTAGE WAVEFORM FOR DELAY TIME



VOLTAGE WAVEFORM FOR  
RISE AND FALL TIMES

NOTES: A.  $C_L = 50$  pF for TLC540 and 100 pF for TLC541

B.  $t_{en} = t_{pZH}$  or  $t_{pZL}$ ,  $t_{dis} = t_{pHZ}$  or  $t_{pLZ}$

C. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.

## TYPES TLC540M, TLC540I, TLC541M, TLC541I 8-BIT ANALOG-TO-DIGITAL PERIPHERALS WITH SERIAL CONTROL AND 11 INPUTS

### principles of operation

The TLC540 and TLC541 are each complete data acquisition systems on a single chip. They include such functions as analog multiplexer, sample-and-hold, 8-bit A/D converter, data and control registers, and control logic. For flexibility and access speed, there are four control inputs [two clocks, chip select ( $\overline{CS}$ ), and address]. These control inputs and a TTL-compatible 3-state output are intended for serial communications with a microprocessor or microcomputer. With judicious interface timing, with TLC540 a conversion can be completed in 10 microseconds, while complete input-conversion-output cycles are being repeated every 14 microseconds. With TLC541 a conversion can be completed in 19 microseconds, while complete input-conversion-output cycles are repeated every 35 microseconds. Furthermore, this fast conversion can be executed on any of 11 inputs or its built-in "self-test," and in any order desired by the controlling processor.

Though they can be operated "tied" together, the System Clock and I/O Clock are normally used independently, with no special phase or speed relationship to be considered. This allows integrated circuit operation to continue independent of serial Input/Output timing, permitting manipulation of the I/O Clock as desired for a wide range of software and hardware needs.

The I/O Clock, Data Input, and Data Output are controlled by  $\overline{CS}$ . It floats the 3-state output and shuts off signals to other control inputs while it is high. This allows any pins except pin 15 to share lines with other integrated circuits. A normal control sequence is as follows: (1)  $\overline{CS}$  goes low; (2) a new positive-logic multiplexer address is clocked in through the address input on the first four I/O Clock rising edges while previous conversion results are brought out on the first seven I/O Clock falling edges. Input and output most-significant bits (MSB) are first, with the output MSB available at the start of the cycle; (3) the on-chip sample-and-hold begins sampling a newly addressed input after the 4th falling edge, and goes into the hold mode on the 8th falling I/O Clock edge just before conversion; (4)  $\overline{CS}$  must then go high or the I/O Clock must remain low for at least 40 system clock cycles to allow conversion. A new address may then be loaded or the previous conversion results read any time  $\overline{CS}$  is brought low, but it should be noted that any pending conversion may stop.

The instant that the TLC540 or TLC541 holds a sample of the analog input, conversion can be determined under software control (or by external logic), by keeping the 8th I/O Clock cycle high. Any output data will have already been shifted out, and TLC540 or TLC541 will continue sampling a new analog input. At the desired time, the I/O Clock signal can then be lowered freezing the voltage and turning off all analog inputs. In this manner, signals can be sampled at precise intervals for a wide range of comparison or processing applications, in much the same manner as a strobe light is used to determine engine speed.

TEXAS  
INSTRUMENTS

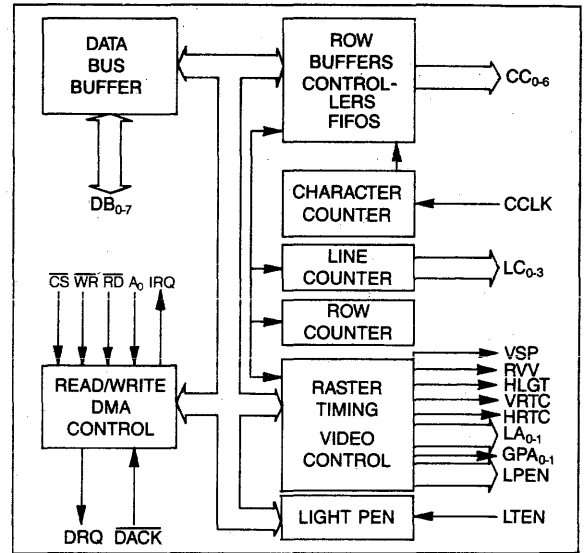
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# NEW PRODUCT BULLETIN

## Programmable CRT Controller WD8275

### FEATURES

- PROGRAMMABLE CHARACTER/SCREEN FORMAT
  - 1-16 LINES PER ROW
  - 1-64 ROWS PER FRAME
  - 1-80 CHARACTERS PER ROW
- SIX FIELD ATTRIBUTES
  - BLINK
  - HIGHLIGHT
  - REVERSE VIDEO
  - UNDERLINE
  - 2-USER PROGRAMMABLE
- ELEVEN CHARACTER ATTRIBUTES
  - RIGHT ANGLE LINE GRAPHIC SYMBOLS
- FOUR CURSOR CONTROLS
  - BLINK/NONBLINK UNDERLINE
  - BLINK/NONBLINK REVERSE VIDEO BLOCK
- LIGHT PEN DETECTION
- REQUIRES +5 VOLT ONLY
- 40 PIN CERDIP AND PLASTIC PACKAGES
- COMPATIBLE WITH INTEL D8275/P8275



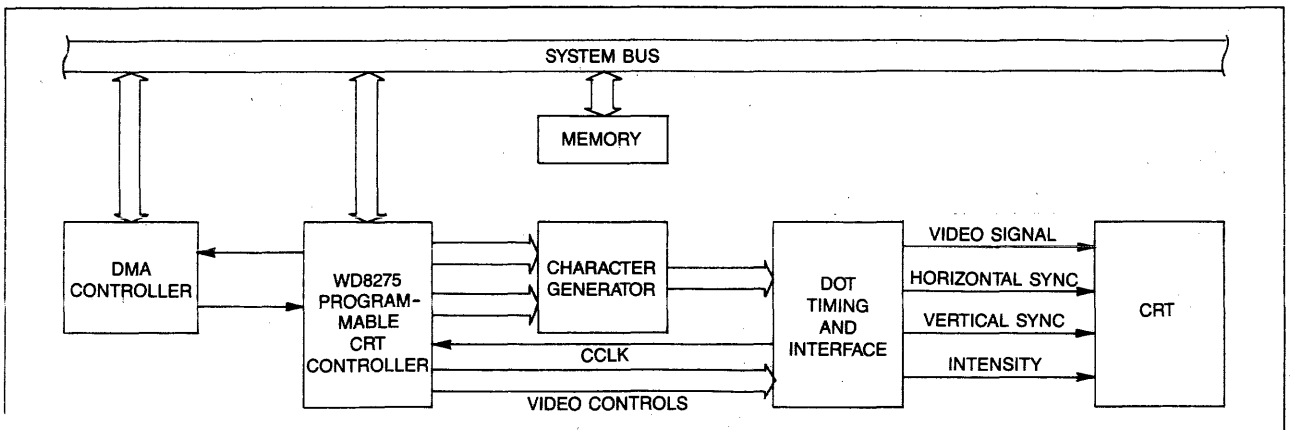
**WD 8275 BLOCK DIAGRAM**

### DESCRIPTION

The WD8275 programmable CRT Controller is a single chip device used to interface raster scan displays with microcomputer systems. Its function is to re-refresh the display by buffering data from main memory and by tracking the display position on the screen. The design flexibility will allow simple interface to raster scan CRT's with a minimum of external hardware/software overhead.

### PROPOSED APPLICATIONS

May be used in a wide variety of terminal applications both remote and resident to the computer system. Additionally, this is an ideal device for Light Pen applications. It is available in 2MHZ and 3MHZ versions (CCLK).



**SIMPLIFIED SYSTEM DIAGRAM**

June, 1983

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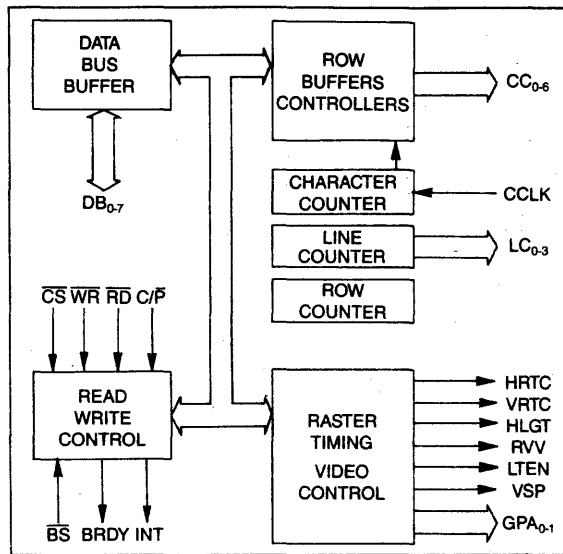
(714) 863-0102, TWX 910-595-1139

# NEW PRODUCT BULLETIN

## Small System CRT Controller WD8276

### FEATURES

- PROGRAMMABLE CHARACTER/SCREEN FORMAT
  - 1-16 LINES PER ROW
  - 1-64 ROWS PER FRAME
  - 1-80 CHARACTERS PER ROW
- SIX FIELD ATTRIBUTES
  - BLINK
  - HIGHLIGHT
  - REVERSE VIDEO
  - UNDERLINE
  - 2-USER PROGRAMMABLE
- FOUR CURSOR CONTROLS
  - BLINK/NONBLINK UNDERLINE
  - BLINK/NONBLINK REVERSE VIDEO BLOCK
- REQUIRES +5 VOLT ONLY
- 40 PIN CERDIP AND PLASTIC PACKAGES
- COMPATIBLE WITH INTEL D8276/P8276



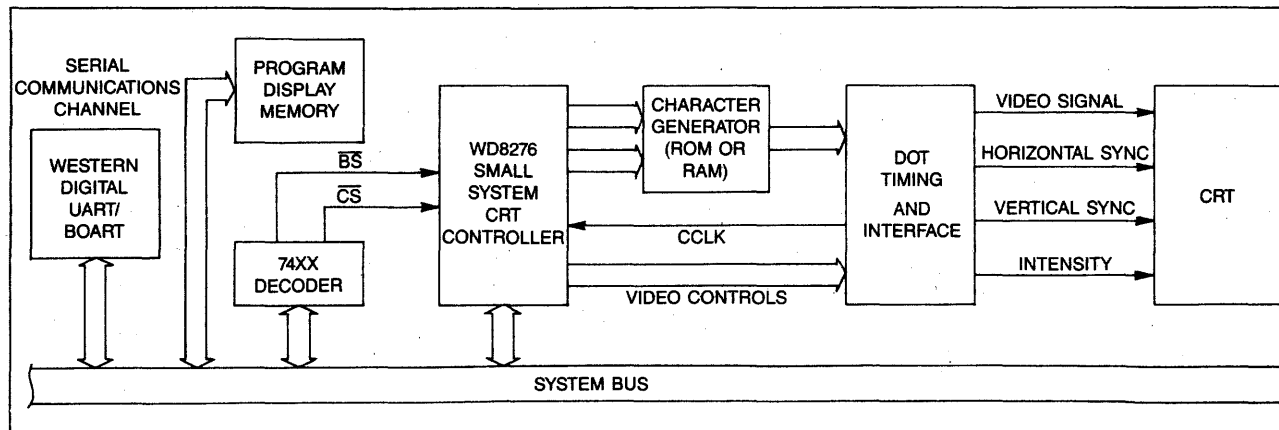
**WD8276 BLOCK DIAGRAM**

### DESCRIPTION

The WD8276 Small System CRT Controller is a single chip device used to interface a raster scan CRT display with a microcomputer in minimum device count systems. It contains the necessary building blocks to refresh the display through character buffering and display position tracking. The flexible design allows simple interface between CRT display and system data bus with a minimum of external hardware.

### PROPOSED APPLICATIONS

An ideal, low cost device for both terminal and small computer system designs. It is available in both 2MHZ and 3MHZ versions (CCLK).



**SIMPLIFIED SYSTEM DIAGRAM**

June, 1983

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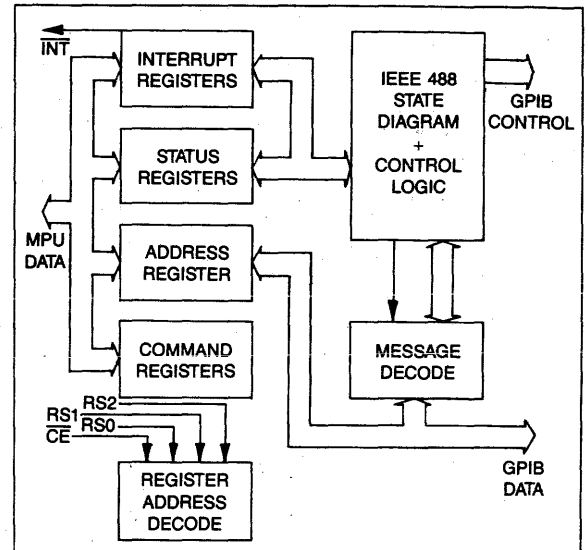
(714) 863-0102, TWX 910-595-1139

# NEW PRODUCT BULLETIN

## General Purpose Interface Bus (GPIB) Controller WD9914

### FEATURES

- IEEE STD 488 (1975/8) COMPATIBLE
- TALKER/LISTENER FUNCTIONS WITH EXTENDED ADDRESSING
- SYSTEM CONTROLLER CAPABILITY
- SERVICE REQUEST
- SERIAL/PARALLEL POLLING
- DEVICE CLEAR/TRIGGER
- DMA FACILITIES
- COMPATIBLE WITH MOST MICROPROCESSORS
- REQUIRES +5VDC ONLY
- 40 PIN CERDIP/PLASTIC PACKAGING
- COMPATIBLE WITH TEXAS INSTRUMENTS TMS9914A



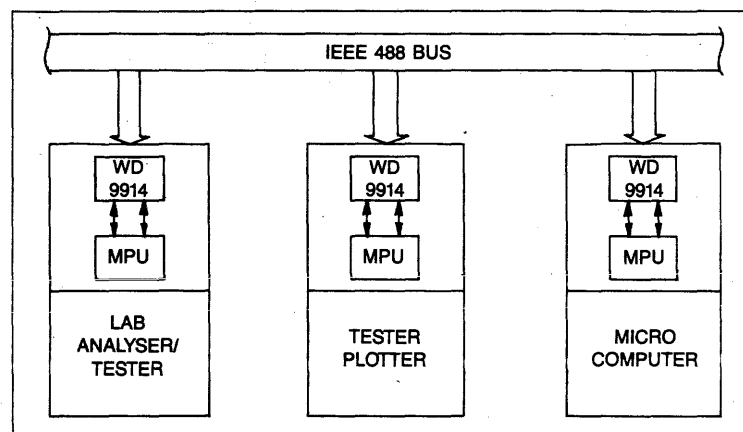
**WD9914 BLOCK DIAGRAM**

### DESCRIPTION

The WD9914 GPIB Controller is a microprocessor controlled device designed to perform the interface function between a microprocessor and the IEEE 488 General Purpose Interface Bus. It automatically handles all Talker, Listener, and Controller operational modes. This relieves the microprocessor from maintaining bus/protocol.

### PROPOSED APPLICATIONS

The WD9914 is used in a variety of instruments interconnected by the GPIB and remotely or automatically programmed and controlled. The following is an example:



**SIMPLIFIED SYSTEM DIAGRAM**

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IRVINE, CALIFORNIA 92714

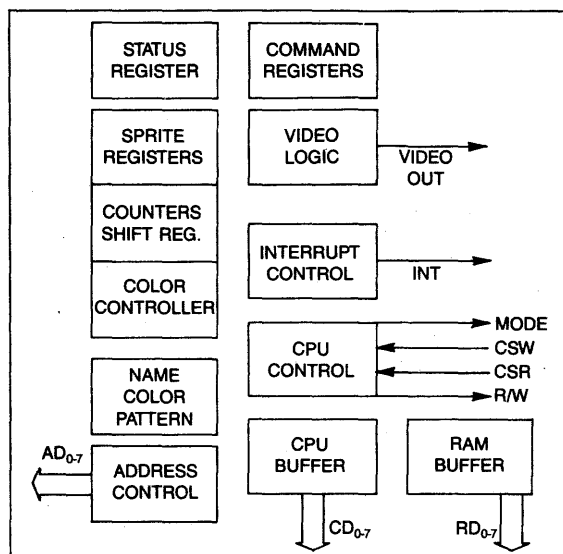
(714) 863-0102, TWX 910-595-1139

# NEW PRODUCT BULLETIN

## Video Display Processor WD9918/9928/9929

### FEATURES

- 32 SPRITE (OBJECT ANIMATION) PLANES
- 15 UNIQUE COLORS
- GENERAL 8 BIT CPU INTERFACE
- DIRECT INTERFACE TO 4K, 18K OR 16K RAMS
- 256 X 192 TV SCREEN RESOLUTION
- COMPOSITE VIDEO OUTPUT (WD9918)
- COLOR DIFFERENCE OUTPUTS (WD9928/9929)
- EUROPEAN PAL (625 LINE) TV (WD9929)
- REQUIRES +5 VOLT ONLY
- 40 PIN CERDIP/PLASTIC PACKAGES
- COMPATIBLE WITH TEXAS INSTRUMENTS (TMS9918A, 9928A, 9929A)



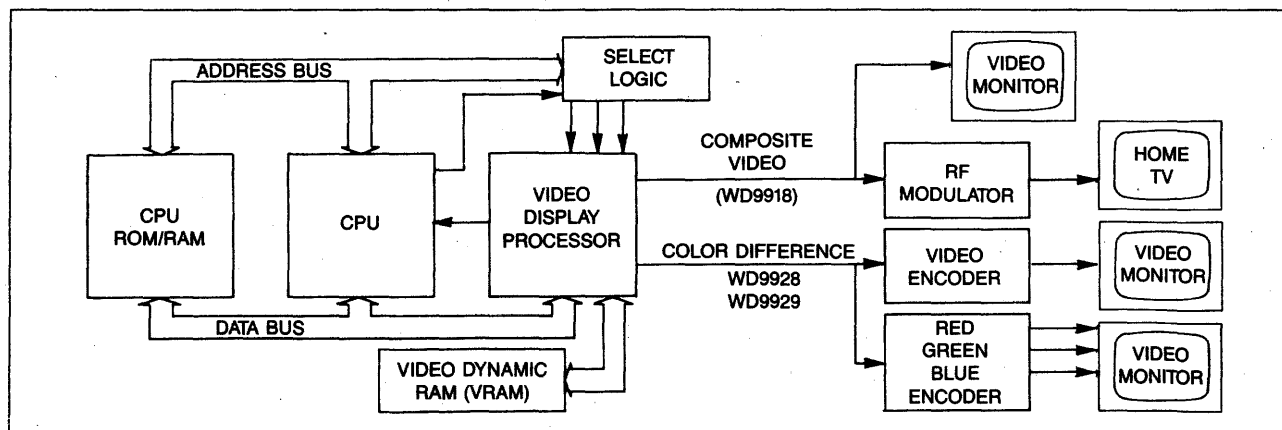
WD9918 BLOCK DIAGRAM

### DESCRIPTION

The Video Display Processor is used to interface a microprocessor and screen refresh memory with a raster scan color CRT display. The device generates all necessary video, control and synchronization signals. Additionally, it controls the storage, retrieval, and refresh of display data in the dynamic screen refresh memory.

### PROPOSED APPLICATIONS

The WD9918/28/29 are used in any video system requiring the display of digital data on a home color television or color monitor. This includes color computer terminals, home computers, process monitoring, and drafting aids.



SIMPLIFIED SYSTEM DIAGRAM

June, 1983

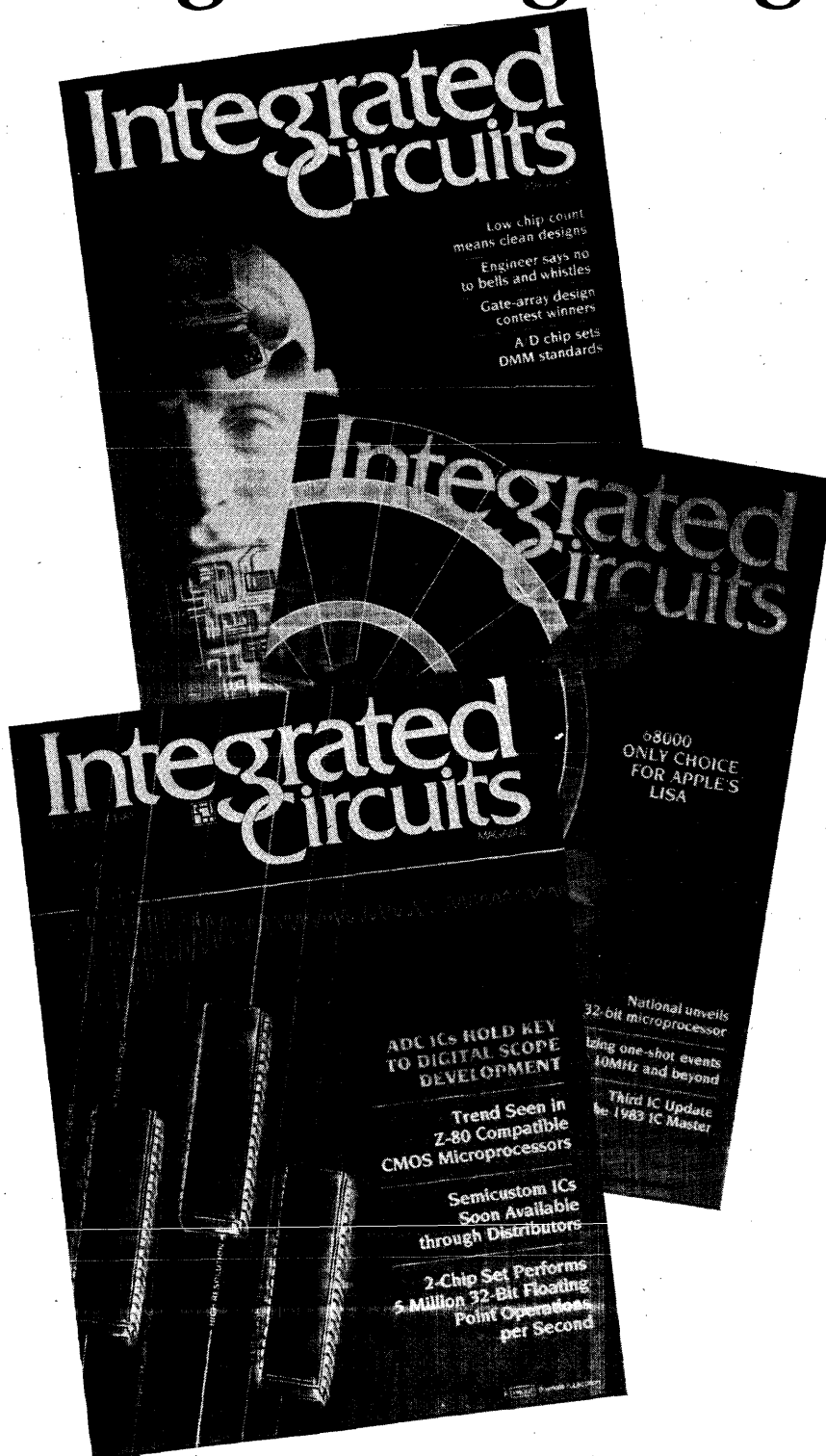
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# INTRODUCTION TO LINEAR

The Master Selection Guide provides sufficient information to make initial product selections. All devices that appear in this section, both in the initial selection guide and the data pages, are included in the indexes. These index listings lead to the page and the line on that page where each device appears.

In the Linear section over 1100 operational amplifiers are covered. Since there are so many devices, the operational amplifier entries have been given special consideration. Separate lists are provided for those which have High Speed, High Voltage capability, Wide bandwidth, etc. Under General Purpose, four amplifier types are listed; these are the ones that the high volume manufacturers indicate are the most widely used; however, quad amplifiers should also be considered when appropriate. If you have located an op amp in a specialized category, you can review its characteristics by finding it in the Part Number or Product Indexes and looking it up in the Operational Amplifier Characteristics and in the Data Sections.

Following the special lists, the "Operational Amplifier Characteristics" listings categorize amplifiers by input parameters. They are arranged in order of increasing offset voltage, bias current, offset current and then voltage drift. The other parameters listed do not affect the sequence in which the devices are presented. The column labeled "Comp" indicates the number of external components normally used for compensation for example "0" means no compensation is required.

Consumer circuits such as audio amplifiers, AM, FM, and TV circuits as well as some digital devices (watches, calculators, etc.) are covered by model number in the Consumer Circuit Section. Linear devices and unusual circuits which do not fit elsewhere are listed under the heading "Other Linear Devices."

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The manufacturers listed above have provided detailed information on their latest and most significant products.

# EINFÜHRUNG LINEAR- SCHALTUNGEN

Der Master Selection Guide für Linearschaltungen enthält alle Informationen, die Sie für die Erstauswahl Ihres Produkts benötigen. Die Bauteile, die in diesem Abschnitt erscheinen, sowohl im Selection Guide als auch auf den Datenblättern, sind in allen Master Indexes enthalten. Diese Register verweisen auf die Seite und Zeile, auf der das entsprechende Bauelement vorkommt.

Der Linear-Teil enthält über 1100 Operationsverstärker. Da es hier sehr viele Bausteine gibt, haben die Eintragungen über Operationsverstärker besondere Beachtung gefunden. Es gibt getrennte Listen für solche mit hoher Geschwindigkeit, hoher Spannung, großer Bandbreite usw. Unter "Allgemeine Verwendung" (General Purpose) finden sich vier Verstärkertypen. Laut Aussage der großen Hersteller sind dies die Gebräuchlichsten. Jedoch sollten auch Vierfachverstärker auf ihre Verwendbarkeit hin untersucht werden. Haben Sie einen Operationsverstärker in einer bestimmten Kategorie gefunden, so schlagen Sie ihn im numerischen Typenverzeichnis oder im Produktverzeichnis nach und überprüfen seine Eigenschaften unter der Rubrik "Operational Amplifier Characteristics" und in den Datenblättern.

Der Abschnitt "Eigenschaften von Operationsverstärkern" (Operational Amplifier Characteristics) bestimmt die Verstärker nach Eingangs-Parametern. Sie sind nach steigender Offsetspannung, Bias-Strom, Offsetstrom und Spannungsdrift angeordnet. Die übrigen aufgeführten Parameter haben keinen Einfluß auf die Reihenfolge, in der die Bauteile genannt werden. Die Spalte "Comp" nennt die Anzahl externer Bausteine, die normalerweise für die Kompensation verwendet werden. "O" z.B. bedeutet, daß keine externe Kompensation erforderlich ist.

Consumer-Schaltkreise wie Audio-Verstärker, AM, FM und TV-Schaltungen sowie einige Digital-Bauteile (Uhren, Rechner usw.) erscheinen nach Typennummer im Abschnitt "Consumer Circuit Section." Solche Linear-Bauelemente und Schaltungen, die in keine entsprechende Rubrik passen, werden unter der Überschrift "Sonstige Linear-Bauelemente" (Other Linear Devices) aufgeführt.

Neu in diesem Jahr sind erweiterte Listen von Sample/ Hold Verstärkern und von festen und einstellbaren Spannungsreferenzen (Fixed and Adjustable Voltage References). Zusätzlich zu der Auflistung unter "Other Linear Devices" finden Sie wichtige Leistungsparameter für diese Produkte am Schluß des Linear-Hauptteils.

# INTRODUCTION AUX SYSTEMES LINEAIRES

Le Guide Général de Sélection fournit suffisamment de renseignements pour permettre des sélections initiales de produits. Tous les appareils mentionnés dans cette Section, à la fois dans le Premier Guide de Sélection et dans les feuilles de données, sont inclus dans tous les index. Ces index indiquent à quelle page et à quelle ligne il a été fait mention de tel ou tel appareil.

La Section "Linéaire" décrit plus de 1100 amplificateurs opérationnels. Du fait du nombre considérable d'appareils entrant dans cette catégorie, une attention toute spéciale a été apportée lors de leur classification. Des listes énumèrent séparément les appareils à grande vitesse, à capacité de haut voltage, à grande plage de fréquence, etc. Dans la rubrique "Applications Générales" (General Purpose), quatre types d'amplificateurs sont cités. Ce sont, d'après les gros fabricants, ceux qui sont le plus couramment usités. Les amplificateurs "QUAD" sont également dignes d'attention dans certains cas. Si vous recherchez un amplificateur opérationnel dans une catégorie bien définie, vous pourrez en obtenir les caractéristiques en consultant les index des Numéros de Pièces ou de Produits, ainsi que la Section "Caractéristiques des Amplificateurs Opérationnels" et les feuilles de données.

La Section "Caractéristiques des Amplificateurs Opérationnels", offre une liste d'amplificateurs classés par paramètre d'entrée. Ces derniers font apparaître dans l'ordre : la tension de compensation, l'ajustement intensité/courant, la compensation du courant, et enfin le changement de tension. Les autres paramètres indiqués n'affectent pas l'ordre dans lequel les appareils sont présentés. La colonne intitulée "COMP" indique le nombre de composants externes également utilisés dans un but de compensation, exemple : "0" signifie qu'aucune compensation n'est nécessaire.

Les circuits de grande consommation tels que les amplificateurs radio AM, FM, les amplificateurs télévision, ainsi que les appareils digitaux (montres, calculatrices, etc.) sont regroupés par numéro de modèle dans la Section "Circuits de Grande Consommation" (Consumer Circuit Section). Les appareils linéaires et autres circuits de moins grande utilisation qui n'entrent pas dans les autres catégories, sont énumérés dans la Section "Autres Appareils Linéaires" (Other Linear Devices).

Un additif nouveau à l'édition de cette année : des références sur les amplificateurs partiels/continus, et avec voltage fixe ou ajustable. En plus des listes fournies dans la Section "Autres Appareils Linéaires", d'autres paramètres de performance relativement importants pour ces produits, apparaissent à la fin du Guide Général de Sélection Linéaire.

# INTRODUCCIÓN A LINEAL

La Guía Maestra de Selección provee suficiente información para hacer selecciones iniciales del producto. Todas las componentes que aparecen en esta sección, ya sea en la guía de selección inicial o en las páginas de datos, están incluidas en todos los otros índices. Estas listas de índices los refiere a la página y línea de aquella página donde se encuentra cada componente.

En la sección bajo Lineal, aparecen más de 1,100 amplificadores operacionales. Debido a que hay tantas componentes, se les ha dado consideración especial a los amplificadores operacionales. Listas apartes son provistas para aquellos con Alta Velocidad, Capacidad de Alto Voltaje, Banda Ancha, etc. Bajo Propósito General, aparecen cuatro tipos de amplificadores, éstos son los que los fabricantes mayores piensan son los más usados; sin embargo, los amplificadores cuadrangulares deben también ser considerados cuando sea necesario. Si usted ha localizado un amp. op. en una categoría especializada, puede usted estudiar sus características encontrándolo en el Índice de Producto o Número de Pieza y buscándolo en la Sección de Datos de las Características de los Amplificadores Operacionales.

Siguiendo las listas especiales, la lista de "Características del Amplificador Operacional" categoriza los amplificadores por parámetros de entrada. Estos están arreglados por orden ascendente de tensión contrapuesta corriente de polarización, corriente contrapuesta, y también deslizamiento de voltaje o tensión. Los otros parámetros mencionados no afectan la secuencia en que las componentes aparecen. La columna titulada "Comp" indica el número de componentes externas que normalmente se emplean para compensación; por ejemplo "O" significa que no requiere compensación.

Circuitos del Consumidor como audioamplificadores, AM, FM, y circuitos de TV así como también algunas componentes digitales (relojes, calculadoras, etc.) aparecen por número de modelo en la Sección de Circuitos del Consumidor. Componentes Lineales y circuitos poco comunes que no corresponden en otras secciones aparecen bajo el título "Otras Componentes Lineales."

Listas expandidas para Amplificadores de Muestra/Retención y Referencias de Voltaje Fijas y Ajustables son nuevas para este año. Además de las listas bajo Otras Componentes Lineales, parámetros de redimiento significantes para estos productos aparecen al final de la Guía Maestra de Selección.

# リニアへの案内

マスターセクションガイドは素子選択にとりかかる充分な資料を描えています。最初のセクションガイド並にデータ掲載ページとこのセクションに記載されている製品はどのインデックにも入っていて、そのページ数と行数がすぐ分ります。

ここでは1,100種以上のオペアンプが収録されています。余りに多品種のため、特別の配慮をしました。それはオペアンプを高速、高耐圧、高域巾他と別々に掲載しました。搬用オペアンプには4種のアンプをのせてあります。大量生産している会社が最も巾広く使用されていると云う品種です。がクオドアンプももし適当なものがあれば考慮すべきかも知れません。

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特殊品種リストに続きオペアンプ特性表があり入出パラメータ毎にオペアンプが分類されています。製品はオフセット電圧、バイアス電流、オフセット電流、電圧ドリフト等小さな順に記載されています。“Comp”と表示されている項目の下の数字は0調整をする為に外付部品が何個必要かを示しています。例えば“0”となっていれば0調整不要を意味します。

民生用製品 - オーディオアンプ、AM、FM、TV素子、ディジタル素子(時計用、電卓用等)は“コンシューマーサーキットセクション”に記載されています。他のどのカテゴリーにも入らないリニアや特異な素子は“アザーリニアデバイス”の見出しに記載されています。

本年新たに加わったものとしてサンプル/ホールドアンプと固定、調整可能なポルティジリファレンスが広くのっています。“アザーリニアデバイス”欄にのっている製品中顕著な性能パラメーターはリニアマスターセクションガイドの最後に掲載されています。

LINEAR-Amplifiers, Special Purpose

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Special Purpose</b>				<b>Floppy Disc Write Amplifier</b>				<b>Instrumentation, Pin Programmable Gain</b>			
<b>AC Amplifier, Quad, Single Supply</b>				<b>XR2247</b> Exar (3371,3372)				AD624A AD			
CA3048 † RCA (3601)				<b>Front End Amplifier (for ultrasonic or infrared remote control systems)</b>				AD624B AD			
CA3052 RCA (3601)				TDA4180 Telefunken				AD624C AD			
<b>AGC Amplifier, 20 Hz to 10 MHz, 6 dB Dynamic Range</b>				<b>Hearing Aid Amplifiers. See Linear-Consumer Circuits</b>				AD624S † AD			
MY108 AnalogSys				<b>Instrumentation Amplifier</b>				<b>Instrumentation (unity gain)</b>			
<b>AGC Generator, for SSB Receivers</b>				INA101SM † Burr-Brown (2851)				3627 Burr-Brown (2851)			
SL621C Plessey				<b>MPINA101A</b> † MicroPwr				<b>Isolation Amplifier</b>			
<b>AGC Generator (voice operated gain adjusting device)</b>				<b>MPINA101B</b> † MicroPwr				MP225A Analogic			
SL620C Plessey				<b>MPINA101C</b> † MicroPwr				MP227 Analogic			
SL6270 Plessey				<b>MPINA101S</b> † MicroPwr				<b>Isolation Amplifier, High Gain</b>			
<b>Amplifier, CATV. See Linear-Consumer Circuits, TV</b>				<b>Instrumentation (commutating auto zero)</b>				MP226 Analogic			
<b>Amplifier for IR Remote Control (1300 dB gain, 8 kHz bandwidth)</b>				ICL7605C Intersil				<b>Isolation Amplifier, Optically Coupled</b>			
U250 Telefunken				ICL7605M † Intersil				3650HG Burr-Brown (2852)			
U3043M Telefunken				ICL7606C Intersil				3652HG Burr-Brown (2852)			
U3083M Telefunken				ICL7606M † Intersil				ISO100AP Burr-Brown (2852)			
<b>Amplifier with Photodiode</b>				<b>Instrumentation, (different input, independent gain adjustment)</b>				<b>Isolation Amplifier, Transformer Coupled</b>			
TFA1001 Siemens				AMP-01 † PMI				3456 Burr-Brown (2851)			
<b>Audio Amplifier with Three NPN Transistors</b>				<b>Instrumentation, (differential input, independent gain adjustment)</b>				3656AG Burr-Brown (2851)			
LM389 National				AD521J AD				<b>Isolation Amplifier, 4 Channel</b>			
<b>Current Amplifier (unity gain, 100 mA output)</b>				AD521K AD				AH427 Analogic			
3329 † Burr-Brown				AD521L AD				<b>Linear-Antilog Amplifier (current in, current out)</b>			
HA-5033 † Harris (3462)				AD521S † AD				SSM2010 SSM			
LH0002 † National				3626A Burr-Brown				SSM2012 SSM			
LH0002C National				3626B Burr-Brown				<b>Linear-Antilog Amplifier, Dual (differential input, separate control inputs)</b>			
<b>Current Amplifier (unity gain, 200 mA output)</b>				3626C Burr-Brown				SSM2000 SSM			
3553 † Burr-Brown (2850,2851)				3629 † Burr-Brown (2851)				SSM2020 SSM			
9963 † OEI				3630A Burr-Brown (2851)				SSM2022 SSM			
<b>Current Amplifier (unity gain, 300 mA output)</b>				3630B Burr-Brown (2851)				<b>Log Amplifier</b>			
MC1438 Motorola				3630C Burr-Brown (2851)				XR7000 Exar (3385)			
MC1538 † Motorola				3630S Burr-Brown (2851)				<b>Log Amplifier, Bipolar (module)</b>			
2003 TeledyneP				INA101 Burr-Brown (2851)				2531A OEI			
2003-01 † TeledyneP				INA104 Burr-Brown				<b>Log/Antilog Amplifier, (antilog)</b>			
<b>Current Amplifier (unity gain, 500 mA output)</b>				MN2200 MicroNet (3042)				ICL8049C Intersil			
9910 † OEI				MN2200H † MicroNet (3042)				2910 † OEI			
9911 † OEI				MN2201 MicroNet				2920 OEI			
<b>Current Amplifier (unity gain, 600 mA output)</b>				MN2201H † MicroNet				SSM2100 SSM			
NA-2630 † Harris (3410)				LH0036 † National				<b>Log/Antilog Amplifier, (log)</b>			
NA-2635 Harris				LH0036C National				4127 Burr-Brown (2852)			
<b>Current Amplifier (wideband, 100 mA output)</b>				LH0037 † National				<b>Log/Log Ratio Amplifier</b>			
HOS-100AH AD (3353)				LH0037C National				LOG100 Burr-Brown (2852)			
HOS-199SH † AD				LH0038 † National				<b>Low Noise, Low Level, Chopper Stabilized Amplifier</b>			
<b>Current Amplifiers. See also Linear-Followers</b>				LH0038C National				MP221 Analogic			
<b>Current Booster, 100 mA Output (for op amp)</b>				AD522A AD				<b>Memory Read Amplifier, Four Head Disk</b>			
2035 † TeledyneP				AD522B AD				TL030 TI			
<b>Differential/Cascode Amplifier (dc to rf)</b>				AD522S † AD				<b>Microphone Amplifier, Electret</b>			
MC3330 Motorola				HC3020 HyComp				PBL3721 RIFA			
CA3028A † RCA (3603)				LM163A † National				<b>Microphone Amplifier (for microphones in telephones)</b>			
CA3028B † RCA (3603)				LM363 National				TDD0246 Telefunken			
CA3053 † RCA (3603)				LM363A † National				<b>Microphone/Headphone Amplifier</b>			
<b>Differential High Speed FET Input Amplifier</b>				<b>Instrumentation, Digitally Programmable Gain</b>				SL6310 Plessey			
MP215 Analogic				PGA100AG Burr-Brown (2851)				<b>Microphone Preamplifier</b>			
<b>Differential Input/Differential Output Amplifier (bandwidth less than 2 MHz)</b>				PGA100BG Burr-Brown (2851)				SSM2015 SSM			
MC1590G † Motorola				LH0084 † National				<b>Mixer (to 200 MHz)</b>			
CA3000 † RCA				LH0084C National				S042 Siemens			
<b>Differential/Video Amplifier</b>				LH0086 † National				<b>Operational Amplifier, Band-Select Switch, AFT Mode Switch (for frequency-synthesizer TV systems)</b>			
CLC103A Comlinear				LH0086C National				CA3166 RCA			
CLC103AM † Comlinear				AM-542MC Datal (2861)							
μPC754 NEC (3547)				AM-542MM † Datal (2861)							
<b>Fiber Optic Receiver Amplifier</b>				AM-542MR Datal (2861)							
LH0082 † National				AM-543MC Datal (2861)							
LH0082C National				AM-543MM † Datal (2861)							
<b>Floppy Disc Read Amplifier System</b>				AM-543MR Datal (2861)							
XB3470A Exar (3372,3373)											
MC3470 Motorola (3061)											
MC3470A Motorola (3061)											
MC3470 TI											

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

## LINEAR-Amplifiers, Special Purpose (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Special Purpose (Cont'd)</b>				<b>RF Amplifiers, Hybrids</b>				<b>Video, IF and RF Amplifiers (Cont'd)</b>			
Operational Amplifier/Comparator with Shutdown Control and Isolated Transistor CA3177 RCA				MHW401 Motorola				TDA3541 Signetics			
Operational Amplifier/Voltage Comparator, Dual LM2924 National				MHW590 Motorola				μA733 † Signetics 115			
LM3924 National				MHW591 Motorola				μA733C Signetics			
LM192 † National				MHW592 Motorola				SG1401 † SiliconG 120			
LM292 National				MHW593 Motorola				SG2401 SiliconG			
LM392 National				MHW601 Motorola				SG3001 † SiliconG			
Operational Amplifier, Dual with Dual Voltage Comparator MC3405 Motorola				MHW602 Motorola				SG3401 SiliconG			
MC3505 † Motorola				MHW603 Motorola				SG733 † SiliconG			
Operational Transconductance Amplifier HA-23080 † Harris (3464)				MHW709 Motorola				SG733C SiliconG			
Operational Transconductance Amplifier, Dual LM13600 Raytheon				MHW710 Motorola				NE592 TI			
LM13600A Raytheon				CA2800 TRW-LSI				TL592 TI			
Power Amplifier, Motor and Actuator Driver, to 12 V, 1.5 A ICH8515I Intersil				CA2810 TRW-LSI				μA733C TI			
ICH8515M † Intersil				CA2820 TRW-LSI				μA733M † TI 125			
Power Amplifier, Motor and Actuator Driver, to 24 V, 3 A ICH8510I Intersil (3479)				CA2830 TRW-LSI				TA7061A Toshiba			
ICH8510M † Intersil (3479)				CA2840 TRW-LSI				Voltage Controlled Amplifier			
ICH8520I Intersil (3479)				CA2842 TRW-LSI				SSM2013 SSM			
ICH8520M † Intersil (3479)				CA2850 TRW-LSI				Wideband Amplifier, Logarithmic Limiting (4 to 500 MHz)			
ICH8530I Intersil (3479)				CA2851 TRW-LSI				SL1531 Plessey			
ICH8530M † Intersil (3479)				CA2870 TRW-LSI				SL1532 Plessey			
Power Transistor Driver/Amplifier, Biases External MOSFET Devices MA500 AnalogSys				CA2871 TRW-LSI				SL531C Plessey			
Preamplifier, for IR Remote Control TDA2320 SGS				CA2875 TRW-LSI				SL532C Plessey			
Preamplifier, for Ultrasonic Remote Control TDA3047 Signetics				CA2876 TRW-LSI				SL532C Plessey			
Preamplifier, Infrared (74 dB gain, 100 kHz bandwidth) TDA4050B Siemens				CA2880 TRW-LSI				Wideband Amplifier with Low Level Video Detection (7 to 200 MHz for log IF amplifiers)			
Preamplifier, Low Noise (35 dB gain, 100 MHz bandwidth) AH0013CA OEI				Signal Sources Switch, (buffer amplifiers with input switches) TDA1028 Signetics				SL1521 † Plessey			
AH0013CB OEI				TDA1029 Signetics				SL1522 Plessey			
AH0013MA † OEI				Thermocouple Amplifier with Cold Junction Compensation AD594A AD (3356)				SL1523 Plessey			
Preamplifier, Precision (to precede operational amplifiers) LM121 † National				AD594C AD (3356)				SL1524 Plessey			
LM121A † National				AD595A AD (3356)				SL1525 Plessey			
LM221 National				AD595C AD (3356)				SL521 Plessey			
LM221A National				Video Buffer Amplifier HA-5033-2 † Harris				SL523 Plessey			
LM321 National				HA-5033-5 Harris				Quad Current Controlled Amplifier SSM2024 SSM 140			
LM321A National				Video, IF and RF Amplifiers CLC100 Comlinear				Hex Analog CMOS Amplifier MA113 AnalogSys			
Preamplifier, Temperature Controlled μA727C Fairchild				CLC102 Comlinear				Two Wire Transmitter (Sends current signal over same two lines from which it is powered.)			
μA727M † Fairchild				CLC104A Comlinear				XTR100AM † Barr-Brown (2851)			
Programmable Channel Op Amp (one of 4 input stages can be connected to single output) HA-2400 † Harris (3394)				CLC104AM † Comlinear				XTR100AP Barr-Brown (2851)			
HA-2404 Harris				CLC200A Comlinear				XTR100BM † Barr-Brown (2851)			
HA-2405 Harris				CLC200AM † Comlinear				XTR100BP Barr-Brown (2851)			
Programmable Gain Amplifier (digitally controlled gain) 3606 Barr-Brown (2851)				CLC220A Comlinear				LH0045 † National			
HS2020 HybridSys				CLC220AM † Comlinear				LH0045C National			
MN2020 MicroNet (3042)				μA733C Fairchild							
Read Amplifier/Preamp (for magnetic tape memory systems) MC3467 Motorola				μA733M † Fairchild							
MC3468 Motorola				μA733 Intersil							
RF Amplifier, Gated MC1445 Motorola				μA733C Intersil							
MC1545 † Motorola				MC1550 † Motorola							
MC1445 TI				MC1733 † Motorola							
			45	NE592 Motorola							
				SE592 † Motorola							
				LM733 † National							
				μPC1651 NEC							
				AN607 Panasonic							
				AN608 Panasonic							
				SL1550 Plessey							
				SL541 Plessey							
				SL550 Plessey							
				SL560 Plessey							
				SL610C Plessey							
				SL611C Plessey							
				SL612C Plessey							
				CA3001 † RCA							
				CA3011 † RCA							
				CA3020A RCA							
				CA3021 † RCA							
				CA3022 † RCA							
				CA3023 † RCA							
				CA3040 † RCA							
				NE592 Signetics (3675)							
				SE592 † Signetics (3675)							
				(Continued)							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR

Master Selection Guide



LINEAR-Arrays

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Transistor Arrays</b>				<b>Dual Transistors, Monolithic NPN Tightly Matched</b>				<b>Quad Darlington Switch</b> (Cont'd)			
<b>Darlington Pair and Two Matched NPN Transistors</b>				(Cont'd)				60			
	CA3018	† RCA	(3601)		MP301	MicroPwr			ULN-2070B	Sprague	
	CA3018A	† RCA	(3601)		MP302	MicroPwr			ULN-2071B	Sprague	
	CA3118	† RCA	(3601)		MP303	MicroPwr			ULN-2074B	Sprague	
	CA3118A	† RCA	(3601)		MP310	MicroPwr			ULN-2075B	Sprague	
<b>Differential Pair and Three NPN Transistors</b>								65			
	μA3086M	Fairchild	5		MP311	MicroPwr			ULN-2076B	Sprague	130
	MC3346	† Motorola			MP312	MicroPwr			ULN-2077B	Sprague	
	MC3386	† Motorola			MP313	MicroPwr			UDN2841	TI	
	LM3045	† National			MP361	MicroPwr			UDN2845	TI	
	LM3046	National			MP362	MicroPwr			ULN2064	TI	
	LM3086	National	10		LM194	† National			ULN2065	TI	135
	LM3146	National			LM394	National					
	SL3045	† Plessey		<b>Dual Transistors, Monolithic NPN, Tightly Matched, Low Noise</b>				70			
	SL3046	Plessey			MAT-02A	† PMI	(3569)	<b>Hex NPN Darlington</b>			
	SL3145	† Plessey			MAT-02B	† PMI	(3569)	LB1274 Sanyo			
	CA3045	† RCA	(3601)	<b>Dual Transistors, Monolithic N-Channel JFET, Tightly Matched</b>				<b>One Darlington PNP Pair, and One Current Mirror Pair with Shared Diode, and Two PNP Transistors</b>			
	CA3046	† RCA	(3601)		MP3954	MicroPwr		CA3084 RCA			
	CA3086	† RCA	(3601)		MP3954A	MicroPwr		<b>Three Differential Amplifiers (NPN)</b>			
	CA3146	RCA	(3601)		MP3955	MicroPwr		MC3350 Motorola			
	CA3146A	RCA	(3601)		MP3956	MicroPwr		ULN-2047A Sprague			
	TBA331	SGS	20		MP3956	MicroPwr		<b>Three NPN and two PNP Transistors</b>			
	TCA671	Siemens			MP3958	MicroPwr		CA3096 † RCA (3601)			
	TCA871	Siemens			MP5911	MicroPwr		CA3096A † RCA (3601)			
	TCA971	Siemens			MP5912	MicroPwr		CA3096C † RCA (3601)			
	TCA991	Siemens			MP5912C	MicroPwr		<b>Four High Current NPN Transistors (core driver)</b>			
	SG3045	SiliconG	25		MP830	MicroPwr		FQ3724 Fairchild			
	SG3046	SiliconG			MP831	MicroPwr		FQ3725 Fairchild			
	SG3086	† SiliconG			MP832	MicroPwr		MHQ4001A Motorola			
	SG3146	SiliconG			MP833	MicroPwr		MHQ4002A Motorola			
	SG3821	SiliconG			MP840	MicroPwr		MHQ4013 Motorola			
	SG3886	SiliconG	30		MP841	MicroPwr		MHQ4014 Motorola			
	ULN-2046A	Sprague			MP843	MicroPwr		MPQ3303 Motorola			
	ULN-2086A	Sprague			MP844	MicroPwr		MPQ3725 Motorola			
	ULS-2045H	† Sprague			MP845	MicroPwr		MPQ3725A Motorola			
<b>Dual Darlington (NPN) and Two NPN Transistors</b>				<b>Dual Transistors, Monolithic PNP, Logarithmic Conformance</b>				90			
	SG3823	SiliconG			MP358	MicroPwr		MPQ4003 Motorola			
<b>Dual Darlington (NPN-PNP quasi complementary)</b>				<b>Dual Transistors, Monolithic PNP, Tightly Matched</b>				95			
	TDA1410	SGS	35		IT130	Intersil		MPQ4004 Motorola			
	TDA1420	SGS			IT130A	Intersil		DH3725C National			
<b>Dual Darlington Switch</b>					IT131	Intersil		DH6376C National			
	ULN-2061M	Sprague			IT132	Intersil		CA1724B † RCA (3601)			
	ULN-2062M	Sprague			MP350	MicroPwr		CA1725B † RCA (3601)			
<b>Dual Differential Amplifiers (NPN), to 120 MHz</b>					MP351	MicroPwr		CA3138 † RCA			
	CA3054	Motorola	40		MP352	MicroPwr		CA3138A † RCA			
	CA3026	RCA		<b>Quad Darlington Switch</b>				100			
	CA3054	RCA			ULN2068	Motorola		<b>Four High Current PNP Transistors (core driver)</b>			
	SG3822	SiliconG			ULN2074	Motorola		FQ3467 Fairchild			
	ULN-2054A	Sprague			L702	SGS		FQ3468 Fairchild			
<b>Dual Differential Amplifiers (NPN), to 500 MHz</b>					SG2064	SiliconG		MHQ3467 Motorola			
	CA3049	† RCA	(3601)		SG2065	SiliconG		DH3467C National			
	CA3102	† RCA	(3601)		SG2066	SiliconG		<b>Five High Current NPN Transistors</b>			
<b>Dual Differential Amplifiers (NPN) with Diode Bias String</b>					SG2067	SiliconG		CA3083 † RCA			
	CA3050	† RCA			SG2068	SiliconG		CA3183 † RCA (3601)			
	CA3051	RCA			SG2069	SiliconG		TDA3083 Signetics			
<b>Dual Transistors (f&gt;5 GHz)</b>					SG2070	SiliconG		SG3083 † SiliconG			
	SL2363C	Plessey			SG2072	SiliconG		SG3183 † SiliconG			
	SL2364C	Plessey			SG2073	SiliconG		SG3183A † SiliconG			
<b>Dual Transistors, Monolithic NPN, Logarithmic Conformance</b>					SG2074	SiliconG		ULN-2083A Sprague			
	MP318	MicroPwr	50		SG2075	SiliconG		ULS-2063H † Sprague			
	MAT-01	† PMI			SG2076	SiliconG		<b>Five High Frequency NPN Transistors, (f&gt;5 GHz)</b>			
	MAT-01A	† PMI			SG2077	SiliconG		SL3127 Plessey			
	MAT-01F	† PMI			UDN-2841B	Sprague		CA3127 RCA (3601)			
	MAT-01G	† PMI			UDN-2845B	Sprague		CA3227 RCA (3601)			
	MAT-01H	PMI	55		ULN-2064B	Sprague		CA3246 RCA (3601)			
<b>Dual Transistors, Monolithic NPN Tightly Matched</b>					ULN-2065B	Sprague		<b>Five High Voltage, High Current NPN Darlington Amplifiers</b>			
	IT120	Intersil			ULN-2066B	Sprague		LB1287 Sanyo			
	IT120A	Intersil			ULN-2067B	Sprague		LB1288 Sanyo			
	IT121	Intersil			ULN-2068B	Sprague		<b>Five High Voltage, High Current NPN Darlington Amplifiers, Source, for Load Connected to Negative Supply</b>			
	IT122	Intersil			ULN-2069B	Sprague		UDN-2956A Sprague			
								UDN-2957A Sprague			
								<b>Five Low-Noise NPN Transistors</b>			
								TDA3310 SGS			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Arrays (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Transistor Arrays (Cont'd)</b>				<b>Seven High Current, High Voltage, NPN Darlington Amplifiers, Open Collector, to 50 V, 500 mA</b>				<b>Eight High Current, High Voltage NPN Darlington Amplifiers, Open Collector, to 50 V, 600 mA</b>			
<b>Six NPN Darlington</b>											
	LB1272	Sanyo			SN75467	TI			ULN-2811A	Sprague	
	LB1273R	Sanyo			ULN2001A	TI			ULN-2812A	Sprague	
<b>Seven High Current Darlington</b>											
	ULN-2031A	Sprague			ULN2002A	TI			ULN-2813A	Sprague	
	ULN-2032A	Sprague			ULN2003A	TI			ULN-2814A	Sprague	
	ULN-2033A	Sprague			ULN2004A	TI			ULN-2815A	Sprague	
<b>Seven High Current, High Voltage, NPN Darlington Amplifiers, Open Collector, to 100 V, 500 mA</b>											
	SN75465	TI			ULN2005A	TI			ULS-2811H	† Sprague	
	SN75468	TI							ULS-2812H	† Sprague	
	SN75469	TI							ULS-2813H	† Sprague	
<b>Seven High Current, High Voltage, NPN Darlington Amplifiers, Open Collector, to 50 V, 500 mA</b>											
	XR2001C	Exar (3383)		<b>Seven High Current, High Voltage NPN Darlington Amplifiers, Open Collector, to 50 V, 600 mA</b>							
	XR2002C	Exar (3383)			ULN-2011A	Sprague			ULS-2814H	† Sprague	
	XR2003C	Exar (3383)			ULN-2012A	Sprague			ULS-2815H	† Sprague	
	XR2004C	Exar (3383)			ULN-2013A	Sprague		<b>Eight High Current, High Voltage NPN Darlington Amplifiers, Open Collector, to 500 mA</b>			
	XR2011C	Exar (3383)			ULN-2014A	Sprague			L601	SGS	
	XR2012C	Exar (3383)			ULN-2015A	Sprague			L602	SGS	
	XR2013C	Exar (3383)			ULS-2011H	† Sprague			L603	SGS	
	XR2014C	Exar (3383)			ULS-2012H	† Sprague			L604	SGS	
	XR2201	Exar		<b>Seven High Current, High Voltage NPN Darlington Amplifiers, Open Collector, to 95 V, 500 mA</b>							
	XR2201M	† Exar			ULN-2021A	Sprague			ULN-2821A	Sprague	
	XR2202	Exar			ULN-2022A	Sprague			ULN-2822A	Sprague	
	XR2202M	† Exar			ULN-2023A	Sprague			ULN-2823A	Sprague	
	XR2203	Exar			ULN-2024A	Sprague			ULN-2824A	Sprague	
	XR2203M	† Exar			ULN-2025A	Sprague			ULN-2825A	Sprague	
	XR2204	Exar			ULS-2021H	† Sprague			ULS-2821H	† Sprague	
	XR2204M	† Exar			ULS-2022H	† Sprague			ULS-2822H	† Sprague	
	9665	Fairchild			ULS-2023H	† Sprague			ULS-2823H	† Sprague	
	9666	Fairchild			ULS-2024H	† Sprague			ULS-2824H	† Sprague	
	9666M	† Fairchild			ULS-2025H	† Sprague			ULS-2825H	† Sprague	
	9667	Fairchild		<b>Seven High Current NPN Transistors, Common Collector</b>							
	9667M	† Fairchild			CA3082	† RCA (3601)		<b>Special Arrays</b>			
	9668	Fairchild			SG3082	SiliconG		<b>CMOS (three p-channel and three n-channel enhancement MOS transistors tested for linear operation)</b>			
	9668M	† Fairchild			ULN-2082A	Sprague			CA3600	† RCA (3601)	
	MC1411	Motorola (3064)		<b>Seven NPN Darlington</b>				<b>Diode Array (one diode quad and two isolated)</b>			
	MC1412	Motorola (3064)			LB1275	Sanyo			CA3019	† RCA (3601)	
	MC1413	Motorola (3064)		<b>Seven NPN Transistors, Common Emitter</b>				<b>Diode Array (10 element)</b>			
	MC1418	Motorola (3064)			CA3081	† RCA (3601)			CA3141	RCA (3601)	
	PBD352301	RIFA			CA3081	Signetics		<b>Diode Array (16 element)</b>			
	PBD352302	RIFA			SG3081	SiliconG			DN803	Panasonic	
	PBD352303	RIFA			ULN-2081A	Sprague		<b>SCR Array (eight SCRs with current-limiting resistors)</b>			
	RBD352301	RIFA		<b>Seven-Stage Driver Array</b>							
	RBD352302	RIFA			LB1260	Sanyo			UTN-2886B	Sprague	
	RBD352303	RIFA			LB1261	Sanyo			UTN-2888A	Sprague	
	L201	SGS			LB1264	Sanyo		<b>Thyristor/Transistor Array (SCR, programmable unijunction transistor, PNP/NPN transistor pair, NPN transistor and zener diode)</b>			
	L202	SGS							CA3087	† RCA (3601)	
	L203	SGS		<b>Eight High Current, High Voltage NPN Darlington Amplifiers, Current Source</b>							
	L204	SGS			UDN-2981A	Sprague		<b>CMOS (three p-channel and three n-channel enhancement MOS transistors tested for linear operation)</b>			
	ULN2001	Signetics			UDN-2982A	Sprague			CA3019	† RCA (3601)	
	ULN2003	Signetics			UDN-2983A	Sprague		<b>Diode Array (10 element)</b>			
	ULN2004	Signetics			UDN-2984A	Sprague			CA3141	RCA (3601)	
	SG2001	SiliconG			UDS-2981H	† Sprague		<b>Diode Array (16 element)</b>			
	SG2002	SiliconG			UDS-2982H	† Sprague			DN803	Panasonic	
	SG2003	SiliconG			UDS-2983H	† Sprague		<b>SCR Array (eight SCRs with current-limiting resistors)</b>			
	SG3851	SiliconG			UDS-2984H	† Sprague			UTN-2886B	Sprague	
	SG3852	SiliconG							UTN-2888A	Sprague	
	SG3853	SiliconG		<b>Eight High Current, High Voltage NPN Darlington Amplifiers, Open Collector, to 50 V, 500 mA</b>							
	ULN-2001A	Sprague			ULN-2801A	Sprague		<b>Thyristor/Transistor Array (SCR, programmable unijunction transistor, PNP/NPN transistor pair, NPN transistor and zener diode)</b>			
	ULN-2002A	Sprague			ULN-2802A	Sprague			CA3087	† RCA (3601)	
	ULN-2003A	Sprague			ULN-2803A	Sprague		<b>CMOS (three p-channel and three n-channel enhancement MOS transistors tested for linear operation)</b>			
	ULN-2004A	Sprague			ULN-2804A	Sprague			CA3019	† RCA (3601)	
	ULN-2005A	Sprague			ULN-2805A	Sprague			CA3141	RCA (3601)	
	ULS-2001H	† Sprague			ULS-2801H	† Sprague		<b>Diode Array (10 element)</b>			
	ULS-2002H	† Sprague			ULS-2802H	† Sprague			CA3141	RCA (3601)	
	ULS-2003H	† Sprague			ULS-2803H	† Sprague		<b>Diode Array (16 element)</b>			
	ULS-2004H	† Sprague			ULS-2804H	† Sprague			DN803	Panasonic	
	ULS-2005H	† Sprague			ULS-2805H	† Sprague		<b>SCR Array (eight SCRs with current-limiting resistors)</b>			
	SN75466	TI							UTN-2886B	Sprague	
									UTN-2888A	Sprague	

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† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR

Master Selection Guide

LINEAR-Comparators

Offset Voltage mV (25°C)	Bias Current (25°C)	Offset Current (25°C)	Response Time ns	Max. Differential Voltage	Gain	Fan Out	Supply Voltage, V	Device	Source	Line
<b>Comparators-Single</b>										
0.25	1.2 $\mu$ A	80 nA	30	—	20K	—	5, -16	RC4805A	Raytheon (3805)	
		1200 nA	80 nA	35 *	—	8000	5 to $\pm$ 18	RM4805A	† Raytheon (3582)	
0.5 *	0.25 nA	0.1 nA	—	30	200K *	5	5, 0 to $\pm$ 15	CMP-05A	† PMI	5
								CMP-05E	PMI	
0.6	1.8 $\mu$ A	150 nA	30	—	20K	—	5, -16	RC4805	Raytheon (3805)	
		1800 nA	150 nA	35 *	—	7000	5 to $\pm$ 18	RM4805	† Raytheon (3582)	
								CMP-05B	† PMI	10
								CMP-05F	PMI	
0.7	0.1 $\mu$ A	40 nA	200	—	200K	5	5	LF2111	† National	
0.8	50 nA	3 nA	270	11	200K	—	5 to $\pm$ 18	CMP-02	† PMI	15
								CMP-02B	PMI	
								CMP-02E	PMI	
	600 nA	25 nA	180	11	200K	—	5 to $\pm$ 18	CMP-01	† PMI	20
								CMP-01B	PMI	
								CMP-01E	PMI	
2.0	10 $\mu$ A	1 $\mu$ A	6.5	6	—	—	-5.2, 6	AM685L	AMD	
			12	6	—	5	-6.5	AM685M	† AMD	
								AM686M	† AMD	25
	15 $\mu$ A	3 $\mu$ A	80	5	12.5K	—	-6, 12	TL810M	† TI	
	20 $\mu$ A	3 $\mu$ A	40	5	40K *	10	-3 to -12, 12	LM106	† National	30
								LM206	National	
			40 *	5	1250	1	-6, 12	$\mu$ A710M	† Fairchild	35
								MC1710	† Motorola	
								LM710	† National	40
								SG710	† SiliconG	
								$\mu$ A710M	† TI	45
								SFC2710	† Thomson-CSF	
	45 $\mu$ A	3 $\mu$ A	40	5	40K	10	-3 to -12, 12	LM106	† TI	
2.8	100 nA	15 nA	270	11	100K	—	5 to $\pm$ 18	CMP-02C	PMI	50
	900 nA	80 nA	180	11	200K	—	5 to $\pm$ 18	CMP-01C	PMI	
3.0	5 $\mu$ A	2 $\mu$ A *	20	5	3K *	2	5, (-6, 5- $\pm$ 15)	LM161	† National	55
								LM261	National	
								AM686C	AMD	60
	10 $\mu$ A	1 $\mu$ A	12	6	—	—	-6.5			
	50 nA	10 nA	200	10	35K	—	$\pm$ 5 to $\pm$ 18	$\mu$ A734M	† Fairchild	
	100 nA	10 nA	165 *	30	200K *	5	5, 0 to $\pm$ 15	LM111	† TI	65
			200 *	30	200K *	5	5, 0 to $\pm$ 15	AD111	† AD	
								AD211	AD	70
								$\mu$ A111M	† Fairchild	
								LM111	† Intersil	75
								LM111	† Motorola	
								LM211	Motorola	80
								LM111	† National	
								LM211	National	85
								LM111	† Raytheon	
								LM111	† Signetics	90
								LM211	Signetics	
								SG111	† SiliconG	95
								SG211	SiliconG	
								SFC2111	† Thomson-CSF	100
								SFC2211	Thomson-CSF	
			250	30	200K *	5	5, 0 to $\pm$ 15	LM111	† AMD	105
								LM211	AMD	
		20 nA	250 *	15	15K	—	5-15, $\pm$ 5- $\pm$ 15	ICL8001M	† Intersil	
3.5	20 $\mu$ A	5 $\mu$ A	80	5	10K	—	-6, 12	TL810C	TI	
4.0	2 $\mu$ A	0.5 $\mu$ A	26	5	5K *	—	5, (-6, 5- $\pm$ 10)	SE527	† Signetics (3821)	
	12 $\mu$ A	3 $\mu$ A	22	5	5K *	—	5, (-6, 5- $\pm$ 10)	SE529	† Signetics (3822)	
	25 pA	50 pA	200	—	200K	2	36 *	LF111	† National	60
								LF211	National	
	50 pA	25 pA	200 *	30	200K *	5	5, 0 to $\pm$ 15	LF111	† AMD	65
								LF211	AMD	
5.0	10 $\mu$ A	2 $\mu$ A	20	5	3K *	2	(-6, 5 to $\pm$ 15)	LM361	National	
		5 $\mu$ A	2.2	5	—	—	5, -5.2	AD9685	AD (3357)	
			2.7	5	—	—	5, -5.2	AD9687	AD (3357)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Comparators (Cont'd)

Offset Voltage mV (25°C)	Bias Current μA (25°C)	Offset Current μA (25°C)	Response Time ns	Max. Differential Voltage	Gain	Fan Out	Supply Voltage, V	Device	Source	Line
<b>Comparators-Single</b>								<b>(Cont'd)</b>		
5.0	20 μA	3 μA	14 *	5	3K *	4	±5	LM160 LM260 LM360	† National National National	(Cont'd)
	25 μA	5 μA	40 40 *	5 5	40K * 1000	10 1	-3 to -12, 12 -6, 12	LM306 μA710C MC1710C LM710C μPC71 SG710C	National Fairchild Motorola National NEC SiliconG	5
	40 μA	5 μA	28 *	5	40K	10	-3 to -12, 12	LM306	TI	10
	75 μA	10 μA	40 *	5	750		-6, 12	TL710M	† TI	
	150 nA	25 nA	200 * 1300 *	10 36	25K 200K *	2	±5 to ±15 2 to 36	μA734C TL331M	Fairchild † TI	
		50 nA	1300 *	36	200K *		2 to 36	TL331C	TI	
	200 nA		50000	32	200K *		2	MK404	AnalogSys	15
	250 nA	50 nA	250 *	15	15K		5-15, ±5-±15	ICL8001C	Intersil	
6.0	2 μA	0.75 μA	26	5	5K *		5, (-6.5-±10)	NE527	Signetics (3621)	
	20 μA	5 μA	22	5	5K *		5, (-6.5-±10)	NE529	Signetics (3622)	
	60 μA	7.5 μA	25 *	5	5K *	2	±4.5 to ±6.5	μA760C μA760M	Fairchild † Fairchild	20
	250 nA	100 nA	250	10	15K		±6 to ±18	AD351J AD351K AD351S	AD AD † AD	
7.5	100 μA	15 μA	40 *	5	700		-6, 12	TL710C	TI	25
	250 nA	50 nA	200 *	30	200K *	5	5, 0 to ±15	AD311 μA311C LM311 LM311 LM311 μPC271 μPC311 LM311 μA311 LM311 SG311 TL311M SFC2311	AD Fairchild Intersil Motorola National NEC NEC Raytheon NEC Signetics SiliconG † TI Thomson-CSF	30 35
	250 pA	50 pA	165 *	30	200K *	5	5, 0 to ±15	LM211 LM311	TI TI	
			250	30	200K *	5	5, 0 to ±15	LM311	AMD	40
		100 pA	—	30	200K *	5	5, 0 to ±15	TL311A	TI	
10	75 pA	150 pA	200	—	200K	2	36 *	LF311	National	
	150 pA	75 pA	200 *	30	200K *	5	5, 0 to ±15	LF311	AMD	
	250 pA	100 pA	—	30	200K *	5	5, 0 to ±15	TL311	TI	
20	500 nA	100 nA	1300 *	Vcc	2K	2	2 to 28	LM3302	TI	45
<b>Comparators-Dual</b>										
2.0	10 μA	1 μA	8 10	6 6			-5.2, 5 -5.2, 5	AM687AM AM687M	† AMD † AMD	
	15 μA	3 μA	80	5	12.5K		-6, 12	TL514M TL820M	† TI † TI	
	20 μA	3 μA	30 * 40 *	5 5	1250 1250	1	-6, 12 -6, 12	LM1514 MC1514	† National † Motorola	50
	100 nA	25 nA	1300 *	36	50K		2 to 36	μA193A LM193A LM193A	† Fairchild † National † Signetics	
		250 nA	50 nA	1300 *	50K		2 to 36	μA293A μA393A	Fairchild Fairchild	55
	250 nA	50 nA	1300 *	36	50K		2 to 36	LM293A LM393A	National National	
3.0	10 μA	1 μA	8 10	6 6			-5.2, 5 -5.2, 5	AM687AL AM687L	AMD AMD	60

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

(Continued)

LINEAR

Master Selection Guide

LINEAR-Comparators (Cont'd)

Offset Voltage mV (25°C)	Bias Current (25°C)	Offset Current (25°C)	Response Time ns	Max. Differential Voltage	Gain	Fan Out	Supply Voltage, V	Device	Source	Line
<b>Comparators-Dual</b>								<b>(Cont'd)</b>		
								(Cont'd)		
3.0	45 $\mu$ A	7 $\mu$ A	40	5	40K *	10	-3 to -12, 12	TL506M	† TI	
	100 nA	10 nA	200 *	30	200K *	5	5, 0 to $\pm$ 15	LH211 LH211 LH2211 LH211 LH211 LH2211	AMD † AMD AMD † Intersil † National National Raytheon	5
			250	30	200K *		5, 0 to $\pm$ 15	AM1500L AM1500M	AMD † AMD	10
3.5	20 $\mu$ A	3 $\mu$ A	80	5	12.5K		-6, 12	TL811M	† TI	
		5 $\mu$ A	80	5	10K		-6, 12	TL514C TL820C	TI TI	
	75 $\mu$ A	10 $\mu$ A	40 *	5	700	1	-6, 12	MC1711	† Motorola	15
					750	1	-6, 12	$\mu$ A711M LM711	† Fairchild † National	
			60	5	750	1	-6, 12	SG711	Silicon6	
4.0	0.5 $\mu$ A	75 nA	80	—	40K	2	$\pm$ 15	LM119	† National	20
	500 nA	75 nA	80 *	5	10K	2	5, 0 to $\pm$ 15	LM119 LM219 LM119 LM219 LM119 LM219 TDC0119	† AMD AMD † National National † Signetics Signetics † Thomson-CSF	25
5.0	5 $\mu$ A *	1 $\mu$ A	2.7 *	30			-5, 2, 5	SP9685 SP9687	Plessey Plessey	
	25 $\mu$ A	5 $\mu$ A	30 *	5	1000	1	-6, 12	LM1414 MC1414	National Motorola	
			40 *	5	1000		-6, 12			
	30 $\mu$ A	5 $\mu$ A	33 *	5	10K		-6, 12	TL811C	TI	30
	100 nA	25 nA	1300 *	36	50K		2 to 36	$\mu$ A193 LM193 LM193 LM193 TDC0193	† Fairchild † National † Signetics † TI † Thomson-CSF	35
	100 $\mu$ A	15 $\mu$ A	40 *	5	700	1	-6, 12	$\mu$ A711C MC1711C LM711C SG711C $\mu$ A711C SFC2711	Fairchild Motorola National Silicon6 TI Thomson-CSF	40
	250 nA	50 nA	1300 *	36	50K		2 to 36	$\mu$ A293 $\mu$ A393 LM293 LM393 LM293 LM293A LM393A LM293 LM393	Fairchild Fairchild National National Signetics Signetics Signetics TI TI	45
					200K *	2	2 to 36/ $\pm$ 1- $\pm$ 18	$\mu$ PC393	NEC	50
6.0	30 pA	15 pA	200 *	36	50K	1	4 to 44	<b>CA3290B</b>	<b>RCA</b>	<b>(3589)</b>
6.5	40 pA	7.5 pA	28 *	5	40K *	10	-3 to -12, 12	TL506C	TI	
7.0	250 nA	50 nA	1500 *	36	25K		2 to 36	$\mu$ A2903 LM2903 LM2903 LM2903 TDF2903	Fairchild National Signetics TI Thomson-CSF	55
	20 $\mu$ A	5 $\mu$ A	18	6	5K *	10	$\pm$ 5	NE521 SE521	Signetics † Signetics	<b>(3620)</b> <b>(526,3620)</b>
			25	6	5K *	10	$\pm$ 5	NE522 SE522	Signetics † Signetics	60
	250 nA	50 nA	200 *	30	200K *	5	5, 0 to $\pm$ 15	LH2311	AMD	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Comparators (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Response Time ns	Max. Differential Voltage	Gain	Fan Out	Supply Voltage, V	Device	Source	Line
<b>Comparators-Dual</b>								<b>(Cont'd)</b>		
7.5	250 nA	50 nA	200 *	30	200K *	5	5, 0 to ± 15	LH2311 LH2311	Intersil National	(Cont'd)
			250	30	200K		5, 0 to ± 15	AM1500C	AMD	
8.0	1 μA	0.2 μA	80 *	5	8K	2	5, 0 to ± 15	LM319 LM319 μPC319 LM319	AMD National NEC Signetics	5
	250 nA	25 nA	600 *	40	—	1	4.5 to 40	MC3324A MC3424A MC3524A	Motorola Motorola Motorola	10
10	40 pA	25 pA	200 *	36	50K	2	4 to 36	CA3290A	RCA (3598)	
	100 nA	25 nA	300 *	38	—	1	4.5 to 40	MC3524	† Motorola	
	250 nA	50 nA	300 *	38	—	1	4.5 to 40	MC3324 MC3424	Motorola Motorola	
20	50 pA	30 pA	200 *	36	25K	2	4 to 36	CA3290	RCA (3598)	15
30	50 pA	100 pA	240	13	20K	1	± 1.5 to ± 7.5	MC14575	Motorola	
			250	13	20K	1	± 1.5 to ± 7.5	MC14575	Motorola	
<b>Comparators-Quad</b>										
0.8	100 nA	25 nA	1300	36	50K	2	2-36/± 1-± 18	CMP-04B CMP-04F	† PMI PMI	
2.0	100 nA	25 nA	1300 *	36	50K	2	2-36/± 1-± 18	LM139A μA139A LM139A LM139A LM139A PM-139A CA139A SG139A LM139A	† AMD † Fairchild † Intersil † Motorola † National † PMI † RCA (3599) † SiliconG † TI	20 25
	250 nA	50 nA	1300 *	36	50K	2	2-36/± 1-± 18	LM239A LM339A μA239A μA339A LM339A LM239A LM339A LM239A LM339A PM-239A PM-339A CA239A CA339A SG239A SG339A	AMD AMD Fairchild Fairchild Intersil Motorola Motorola National National PMI PMI RCA (3599) RCA (3599) SiliconG SiliconG	30 35 40
2.0 *	20 μA	1 μA *	55	5	1.2K *	10	± 5	MC3430 MC3431	Motorola Motorola	45
			65	5	1.2K *	10	± 5	MC3432 MC3433	Motorola Motorola	
2.5	5 nA	0.5 nA	10 ms	36	25K		2 to 36	L161	Siliconix (3083)	
3.0	75 nA	25 nA	130 *	15	400K *		5 to ± 15	HA-4900-2	† Harris (3457)	
5.0	2.5 nA *	0.5 nA *	8000	36	500K	2	2-36/± 1.5-± 18	LP339	National	50
	100 nA	25 nA	1300 *	36	200K *	2	2-36/± 1, ± 18	LM139 μA139 LM139 LM139 LM139 PM-139 LM139 CA139 LM139A SG139 LM139 TDC0139	† AMD † Fairchild † Intersil † Motorola † National † PMI † Raytheon † RCA (3599) † Signetics (526) † SiliconG † TI † Thomson-CSF	55 60

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR

Master Selection Guide

LINEAR-Comparators (Cont'd)

Offset Voltage mV (25°C)	Bias Current (25°C)	Offset Current (25°C)	Response Time ns	Max. Differential Voltage	Gain	Fan Out	Supply Voltage, V	Device	Source	Line
<b>Comparators-Quad</b>								<b>(Cont'd)</b>		
5.0								(Cont'd)		
	150 nA	35 nA	200	15	400K *	—	5 to ± 15	HA-4902-2	† Harris	
	250 nA	50 nA	1300 *	15	200K *	1	3-15	MM54C909	† National	
				36	200K	1	1 to 18	MB4204	Fujitsu	
					200K *	2	2-36/± 1, ± 18	LM239	AMD	5
								LM339	AMD	
								μA239	Fairchild	
								μA339	Fairchild	
								LM339	Intersil	10
								LM239	Motorola	
								LM339	Motorola	
								LM239	National	
								LM339	National	
								μPC177	NEC	15
								μPC339	NEC	
								AM6912	Panasonic (3557)	
								PM-239	PMI	
								PM-339	PMI	
								LM239	Raytheon	20
								LM339	Raytheon	
								CA239	RCA (3599)	
								CA339	RCA (3599)	
								LM239A	Signetics	
								LM339	Signetics	
								SG239	SiliconG	25
								SG339	SiliconG	
								LM239	TI	
								LM339	TI	
5.0 (2 operational amplifiers, 2 comparators)										
	500 nA	75 nA	1300 *	36	200K *		3-36	MC3505	† Motorola	
7.0										
	250 nA	50 nA	1300 *	36	50K	2	2-36/± 1, ± 18	LM2901	Motorola	30
								LM2901	National	
								μPC2901	NEC	
								LM2901	Raytheon	
								LM2901	Signetics	35
								LM2901	TI	
								TDF2901	Thomson-CSF	
7.5										
	150 nA	50 nA	130 *	15	400K *		5 to ± 15	HA-4905-5	Harris	
9.0 (2 operational amplifiers, 2 comparators)										
	10	500 nA	1300 *	36			3-36	MC3405	Motorola	
20										
	500 nA	3 nA *	2000 *	Vcc	2K	1	2 to 28	MC3302	Motorola	40
								MC3302	Signetics	
								SG3302	SiliconG	
								TDF3302	Thomson-CSF	
		100 nA	1300 *	Vcc	2K	2	5 to ± 15	LM3302	National	
30 (2 operational amplifiers, 2 comparators)										
	50 pA	100 pA	250	13	20K	1	± 1.5 to ± 7.5	MC14574	Motorola	
<b>Comparators-Hex</b>										
5										
	100 nA	25 nA	1300	± 36	200K	4	2-36/± 1, ± 18	TL336C	TI	45
	250 nA	25 nA	1300	± 36	200K	4	3-36/± 1, ± 18	TL336M	† TI	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

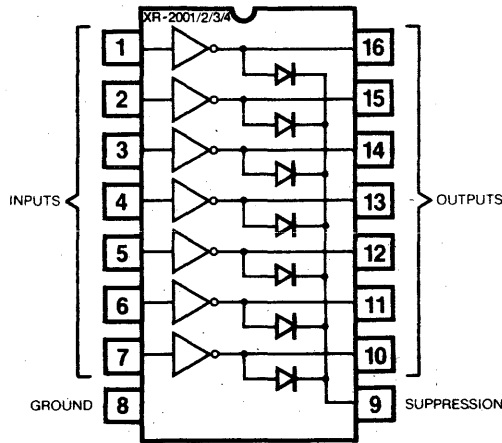
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## High Current Drivers

### XR-2001/2002/2003/2004 HIGH-VOLTAGE, HIGH-CURRENT DARLINGTON TRANSISTOR ARRAYS

The XR-2001/2002/2003/2004 are high-voltage, high-current Darlington transistor arrays, consisting of seven silicon npn Darlington pairs on a common monolithic substrate. All units feature open-collector outputs and integral protection diodes for driving inductive loads. Peak inrush currents of up to 600 mA are allowed, making the arrays ideal for driving tungsten filament lamps. The outputs may be paralleled to achieve high-load capability, although each driver has a maximum continuous collector-current rating of 500 mA. The arrays are directly price competitive with discrete transistor alternatives.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Peak Inrush Current Capability of 600 mA
- Internal Protection Diodes for Driving Inductive Loads
- Excellent Noise Immunity
- Direct Compatibility with Most Logic Families
- Opposing Pin Configuration Eases Circuit Board Layout

#### APPLICATIONS

- Relay Drive
- High-Current Logic Driver

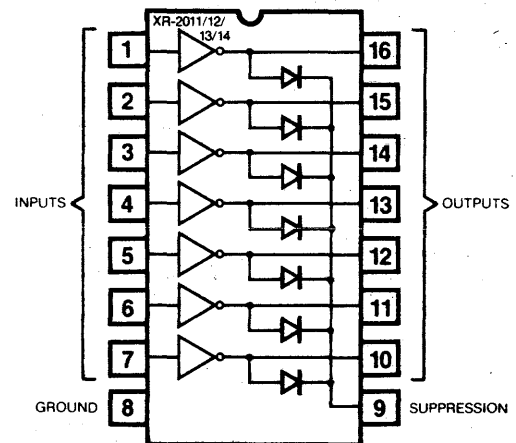
#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2001CN	Ceramic	0°C to +70°C
XR-2002CN	Ceramic	0°C to +70°C
XR-2003CN	Ceramic	0°C to +70°C
XR-2004CN	Ceramic	0°C to +70°C

### XR-2011/2012/2013/2014 HIGH-VOLTAGE, HIGH-CURRENT DARLINGTON TRANSISTOR ARRAYS

The XR-2011/2012/2013/2014 are high-voltage, high-current Darlington transistor arrays, consisting of seven silicon npn Darlington pairs on a common monolithic substrate. All units feature open-collector outputs and integral protection diodes for driving inductive loads. Peak inrush currents of up to 750 mA are allowed, making the arrays ideal for driving tungsten filament lamps. The outputs may be paralleled to achieve high-load capability, although each driver has a maximum continuous collector-current rating of 600 mA. The arrays are price competitive with discrete transistor alternatives.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Peak Inrush Current Capability of 750 mA
- Internal Protection Diodes for Driving Inductive Loads
- Excellent Noise Immunity
- Direct Compatibility with Most Logic Families
- Opposing Pin Configuration Eases Circuit Board Layout

#### APPLICATIONS

- Relay Drive
- High-Current Logic Driver

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2011CN	Ceramic	0°C to +70°C
XR-2012CN	Ceramic	0°C to +70°C
XR-2013CN	Ceramic	0°C to +70°C
XR-2014CN	Ceramic	0°C to +70°C





## Full Custom Development

Exar offers a complete design and production capability for full custom IC development, using Exar's bipolar, I<sup>2</sup>L, and CMOS technologies. This provides an excellent complement to Exar's unique semi-custom capability.

Exar's full custom IC development and production capabilities offer complete flexibility to meet changing customer needs or design problems. We can develop a complete custom IC starting from your black box specifications, or reduce your working breadboard prototype to a monolithic chip. Alternately, if you have the facilities and resources to do the IC design and layout, Exar will provide you with the device characteristics and IC layout rules for the particular process suitable to your design, and review your IC layout for you. Then, Exar can generate the IC tooling and fabricate your IC prototypes for you.

### WAFER FOUNDARY SERVICE

Exar's bipolar process technology is compatible with the manufacturing processes available from many of the other IC manufacturers. Thus, if you have developed a set of IC tooling with another manufacturer and would like an alternate (or substitute) supplier for your custom IC product, in most cases the existing IC tooling will be directly compatible with Exar's technology.

Exar's Engineering department has two custom IC design groups dedicated to the development of linear and digital custom LSI. We pride ourselves in our flexibility and quick response to your needs.

### CONVERTING SEMI-CUSTOM TO FULL CUSTOM

Exar offers the unique ability to start a program using a combination of semi-custom bipolar and/or I<sup>2</sup>L, and CMOS arrays, during the early phases of a customer's product, taking full advantage of the low tooling cost and short development cycle. As a customer's product matures, and its market expands, resulting in higher volume production run rates, Exar can convert the multiple semi-custom chip approach into a single custom IC, achieving a cost reduction, and in many cases a performance improvement. The significant advantage of this type of program is that the risk associated with a custom development is greatly reduced; the IC design approach has been proven, production "bugs" are out of your product, and your production line continues to flow during the full custom chip development. Once the custom chip is completely characterized and found acceptable, the semi-custom IC system in your product can be phased-out while the full custom IC is being phased-in.

Exar is the only company that can offer you the advantages of semi-custom and full custom bipolar design programs, because of our complete in-house semiconductor

manufacturing capability.

### YOUR FIRST STEP

The following technical data package is required in order for Exar to provide you with a firm quotation for your full custom development program:

1. Circuit block diagram with sub-blocks (as necessary).
2. Circuit schematic or logic diagram.
3. Description of circuit operation, and pertinent application information.
4. Preliminary or objective device specification indicating min/max conditions, and limits for the critical parameters (i.e., input/output voltage and current levels, operating frequency, timing diagrams, input/output impedances, power dissipation, etc.).
5. Production requirements and the desired development timetable.

### TYPICAL FLOW FOR FULL CUSTOM DEVELOPMENT

◇  
TECHNICAL FEASIBILITY AND COST PROPOSAL

◇  
CUSTOMER/EXAR INITIAL DESIGN REVIEW

◇  
BREADBOARD ANALYSIS

◇  
COMPUTER SIMULATION FOR WORST CASE DESIGN ANALYSIS

◇  
CUSTOMER/EXAR PROPOSED INTEGRATION DESIGN REVIEW

◇  
CHIP ARCHITECTURE DEFINITION

◇  
CHIP DIGITIZING

◇  
PATTERN GENERATION PHOTOMASKING

◇  
PROTOTYPE WAFER FABRICATION

◇  
PROTOTYPE ASSEMBLY

◇  
PROTOTYPE EVALUATION

**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Voltage Between V and V- pins	40V	Maximum Junction Temperature	200°C
Input Voltage	Equal to Supplies	Operating Temperature Range HA-5033-2	-55°C ≤ TA ≤ +125°C
Output Current	±200mA	HA-5033-5	0°C ≤ TA ≤ +75°C
Internal Power Dissipation (Note 2)		Storage Temperature Range	65°C ≤ TA ≤ 150°C
TO-8 (+25°C)	1.75W		
Mini-dip (+25°C)	1.95W		

**ELECTRICAL CHARACTERISTICS** VSUPPLY = ±12V, RS = 50Ω, RL = 100Ω, CL = 10pf, unless otherwise specified.

PARAMETER	TEMP	-55°C TO +125°C			0°C TO +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		5	15		5	15	mV
	Full		7	25		7	25	mV
Average Offset Voltage Drift	Full		33			33		μV/°C
Bias Current	+25°C		10	35		10	35	μA
	Full		20	50		20	50	μA
Input Resistance	+25°C		1.5			1.5		MΩ
Input Capacitance	+25°C		1.6			1.6		pF
Input Noise Voltage(10Hz-10MHz)			20			20		μVp-p
<b>TRANSFER CHARACTERISTICS</b>								
Voltage Gain RL = 100Ω	+25°C	.93			.93			V/V
RL = 1KΩ	+25°C		.99			.99		V/V
RL = 100Ω	Full	.92			.92			V/V
-3db Bandwidth	+25°C		250			250		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing								V
RL = 100Ω	Full		±10			±10		V
RL = 1KΩ(Note 3)	Full		±11			±11		V
Output Current	+25°C		±100			±100		mA
Output Resistance	+25°C		5			5		Ω
Full Power Bandwidth (Note 4)	+25°C		80			80		MHz
<b>TRANSIENT RESPONSE</b>								
Rise Time	+25°C		3			3		ns
Propagation Delay	+25°C		1			1		ns
Overshoot	+25°C		9			9		%
Slew Rate (Note 3)	+25°C	1.0	1.3		1.0	1.3		V/ns
Settling Time to .1%	+25°C		50			50		ns
Differential Phase Error	+25°C		.1			.1		degrees
Differential Gain Error	+25°C		.1			.1		%
<b>POWER REQUIREMENTS</b>								
Supply Current	+25°C		21	25		21	25	mA
	Full		21	30		21	30	mA
Power Supply Rejection Ratio	Full	54			54			db
Harmonic Distortion	25°C		<0.1			<0.1		%

**NOTES:**

1. Absolute maximum ratings are limiting values, applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.

2. TO-8 θJA = 99°C/W. θJC = 31°C/W; mini-dip θJA = 90°C/W, θJC = 27°C/W

3. VS = ±15V, RL = 1KΩ

4. VIN = 1 VRMS

# ROM Selection Guide

Commercial:  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$

Part Number	Organization	Access Time (ns) Max.	Maximum Current (mA)		Power Supply (Volts)	Number of Pins	Package Type (Note 1)	Compatible EPROM/PROM
			Operating	Standby				
SY3308	1024 x 8	70	120	—	+5	24	C, D, P, K	82S181
SY3308R	1024 x 8	35 <sup>[3]</sup>	130	—	+5	24	C, D, P, K	27S35
SY2316B	2048 x 8	450	98	—	+5	24	C, D, P	2716
SY2316B-2	2048 x 8	200	98	—	+5	24	C, D, P	2716
SY2316B-3	2048 x 8	300	98	—	+5	24	C, D, P	2716
SY3316	2048 x 8	80	120	—	+5	24	C, D, P, K	82S191
SY3316A	2048 x 8	80	120	20	+5	24	C, D, P, K	82S191
SY3316R	2048 x 8	35 <sup>[3]</sup>	130	—	+5	24	C, D, P, K	27S45
SY2332	4096 x 8	450	100	—	+5	24	C, D, P	TMS2532
SY2332-2	4096 x 8	200	100	—	+5	24	C, D, P	TMS2532
SY2332-3	4096 x 8	300	100	—	+5	24	C, D, P	TMS2532
SY2333	4096 x 8	450	100	—	+5	24	C, D, P	2732/A
SY2333-2	4096 x 8	200	100	—	+5	24	C, D, P	2732/A
SY2333-3	4096 x 8	300	100	—	+5	24	C, D, P	2732/A
SY2364	8192 x 8	450	100	—	+5	24	C, D, P	TMS2564
SY2364-2	8192 x 8	200	100	—	+5	24	C, D, P	TMS2564
SY2364-3	8192 x 8	300	100	—	+5	24	C, D, P	TMS2564
SY2364A	8192 x 8	450	100	12	+5	24	C, D, P	TMS2564
SY2364A-2	8192 x 8	200	100	12	+5	24	C, D, P	TMS2564
SY2364A-3	8192 x 8	300	100	12	+5	24	C, D, P	TMS2564
SY2365	8192 x 8	450	100	—	+5	28	C, D, P	2764
SY2365-2	8192 x 8	200	100	—	+5	28	C, D, P	2764
SY2365-3	8192 x 8	300	100	—	+5	28	C, D, P	2764
SY2365A	8192 x 8	450	100	12	+5	28	C, D, P	2764
SY2365A-2	8192 x 8	200	100	12	+5	28	C, D, P	2764
SY2365A-3	8192 x 8	300	100	12	+5	28	C, D, P	2764
SY23128-2	16,384 x 8	200	100	—	+5	28	C, D, P	27128
SY23128-3	16,384 x 8	300	100	—	+5	28	C, D, P	27128
SY23128	16,384 x 8	450	100	—	+5	28	C, D, P	27128
SY23128A-2	16,384 x 8	200	100	10	+5	28	C, D, P	27128
SY23128A-3	16,384 x 8	300	100	10	+5	28	C, D, P	27128
SY23128A	16,384 x 8	450	100	10	+5	28	C, D, P	27128
SY23130	16,384 x 8	450	100	—	+5	28	C, D, P	—
SY23130-2	16,384 x 8	200	100	—	+5	28	C, D, P	—
SY23130-3	16,384 x 8	300	100	—	+5	28	C, D, P	—
SY23130A	16,384 x 8	450	100	10	+5	28	C, D, P	—
SY23130A-2	16,384 x 8	200	100	10	+5	28	C, D, P	—
SY23130A-3	16,384 x 8	300	100	10	+5	28	C, D, P	—
SY23256-2	32,768 x 8	200	100	—	+5	28	C, D, P	27256
SY23256-3	32,768 x 8	300	100	—	+5	28	C, D, P	27256
SY23256	32,768 x 8	450	100	—	+5	28	C, D, P	27256
SY23256A-2	32,768 x 8	200	100	10	+5	28	C, D, P	27256
SY23256A-3	32,768 x 8	300	100	10	+5	28	C, D, P	27256
SY23256A	32,768 x 8	450	100	10	+5	28	C, D, P	27256
SY6364 <sup>[2]</sup>	8192 x 8	200	70	1mA/10 $\mu$ A <sup>[4]</sup>	+5	24	C, D, P	
SY6365 <sup>[2]</sup>	8192 x 8	200	70	1mA/10 $\mu$ A <sup>[4]</sup>	+5	28	C, D, P	

**NOTES:**

1. C = Ceramic, D = Cerdip, P = Plastic, K = Leadless Chip Carrier.
2. Preliminary information.
3. Effective Access Time ( $t_{CPA}$ ).
4.  $\overline{CE}$  @ 2V/ $\overline{CE}$  @  $V_{CC}$ .

# MANUFACTURERS & DISTRIBUTORS DIRECTORY

Intersil (Cont)			
CO	<b>Englewood</b> Kierulff Elctrns., 303-790-4444	MO	<b>St. Louis</b> Arrow Elctrns., 314-567-6888
CO	<b>Thornton</b> Wyle Distribution Grp., 303-424-1985	NH	<b>Manchester</b> Arrow Elctrns., 603-668-6968
CO	<b>Wheatridge</b> Bell Inds., 303-424-1985	NH	<b>Manchester</b> Schweber Electronics, 603-625-2250
CT	<b>Danbury</b> Schweber Elctrns., 203-792-3742	NJ	<b>Evesham</b> Arrow Elctrns., 609-596-8000
CT	<b>Wallingford</b> Arrow Elctrns., 203-265-7741	NJ	<b>Fairfield</b> Arrow Elctrns., 201-575-5300
CT	<b>Wallingford</b> Kierulff Electronics, 203-265-1115	NJ	<b>Fairfield</b> Schweber Elctrns., 201-227-7880
FL	<b>Altamonte Springs</b> Schweber Elctrns., 305-331-7555	NM	<b>Albuquerque</b> Alliance Elctrns., Inc., 505-292-3360
FL	<b>Fl. Landersdale</b> Arrow Elctrns., 305-776-7790	NM	<b>Albuquerque</b> Arrow Elctrns., 505-243-4566
FL	<b>Fl. Landersdale</b> Kierulff Electronics, 305-486-4004	NM	<b>Albuquerque</b> Bell Inds., Century Elctrns. Div., 505-292-2700
FL	<b>Hollywood</b> Schweber Elctrns., 305-921-0301	NY	<b>Binghamton</b> Pioneer Harvey Elec., 607-748-8211
FL	<b>Palm Bay</b> Arrow Elctrns., 305-725-1480	NY	<b>Buffalo</b> Summit Distrs., Inc., 716-887-2800
FL	<b>St. Petersburg</b> Kierulff Electronics, 813-576-1966	NY	<b>Fairport</b> Pioneer Harvey Elec., 716-381-7070
GA	<b>Marietta</b> Arrow Elctrns., 404-449-8252	NY	<b>Hempstead</b> Arrow Elctrns., 516-231-1000
GA	<b>Marietta</b> Schweber Elctrns., 404-449-9170	NY	<b>Liverpool</b> Arrow Elctrns., 315-652-1000
IL	<b>Chicago</b> Newark Elctrns., 312-638-4411	NY	<b>Melville</b> Arrow Elctrns., 516-391-1609
IL	<b>Elk Grove Village</b> Kierulff Elctrns., 312-640-0200	NY	<b>Rochester</b> Arrow Elctrns., 716-275-0300
IL	<b>Elk Grove Village</b> Schweber Elctrns., 312-364-3774	NY	<b>Rochester</b> Schweber Elctrns., 716-424-2222
IL	<b>Schaumburg</b> Arrow Elctrns., 312-397-3440	NY	<b>Westbury</b> Schweber Elctrns., 516-334-7474
IN	<b>Indianapolis</b> Advent Elctrns., Inc., 317-872-4910	NC	<b>Raleigh</b> Arrow Elctrns., 919-876-3132
IN	<b>Indianapolis</b> Arrow Elctrns., 317-243-9353	NC	<b>Raleigh</b> RESCO, 919-781-5700
IA	<b>Cedar Rapids</b> Advent Elctrns., 319-395-0197	NC	<b>Raleigh</b> Schweber Electronics, 919-876-0000
IA	<b>Cedar Rapids</b> Arrow Electronics, 319-373-7230	NC	<b>Winston-Salem</b> Arrow Elctrns., 919-725-8711
IA	<b>Cedar Rapids</b> Schweber Elctrns., 319-373-1417	OH	<b>Beachwood</b> Schweber Elctrns., 216-464-2970
KS	<b>Overland Park</b> Schweber Electronics, 913-492-2922	OH	<b>Bedford Heights</b> Kierulff Electronics, 216-587-6558
MD	<b>Baltimore</b> Arrow Elctrns., 301-247-5200	OH	<b>Centerville</b> Arrow Elctrns., 513-435-5563
MD	<b>Baltimore</b> Kierulff Electronics, 301-247-5020	OH	<b>Centerville</b> Schweber Electronics, 513-439-1800
MD	<b>Beltsville</b> Schweber Elctrns., 301-840-5900	OH	<b>Solon</b> Arrow Elctrns., 216-248-3990
MA	<b>Bedford</b> Schweber Elctrns., 617-275-5100	OK	<b>Tulsa</b> Arrow Electronics, 918-665-7700
MA	<b>Billerica</b> Kierulff Elctrns., 617-667-8331	OK	<b>Tulsa</b> Kierulff Electronics, 918-252-7537
MA	<b>Woburn</b> Arrow Elctrns., 617-933-8130	OK	<b>Tulsa</b> Schweber Electronics, 918-622-8000
MI	<b>Ann Arbor</b> Arrow Elctrns., 313-971-8220	OR	<b>Hillsboro</b> Wyle Distribution, 503-640-6000
MI	<b>Grand Rapids</b> Arrow Electronics, 616-243-0912	OR	<b>Lake Oswego</b> Moore Electronics, Inc., 503-684-3131
MI	<b>Livonia</b> Schweber Elctrns., 313-525-8100	OR	<b>Portland</b> Kierulff Elctrns., 503-641-9150
MI	<b>Eden Prairie</b> Schweber Elctrns., 612-941-5280	PA	<b>Hershey</b> Schweber Elctrns., 215-441-0600
MI	<b>Edina</b> Arrow Elctrns., 612-830-1819	PA	<b>Wheatsville</b> Arrow Elctrns., 412-856-7000
MI	<b>Edina</b> Kierulff Elctrns., 612-941-7500	PA	<b>Pittsburgh</b> Schweber Elctrns., 412-782-1600
MO	<b>Earth City</b> Schweber Electronics, 314-739-0526	TX	<b>Austin</b> Arrow Elctrns., 512-835-4180
MO	<b>Kansas City</b> LCOMP, 816-221-2400	TX	<b>Austin</b> Kierulff Electronics, 512-835-2090
MO	<b>Maryland Heights</b> LCOMP, 314-291-6200	TX	<b>Austin</b> Schweber Electronics, 512-458-8253
		TX	<b>Dallas</b> Arrow Elctrns., 214-386-7500
		TX	<b>Dallas</b> Kierulff Electronics, 214-343-2400
		TX	<b>Dallas</b> Schweber Elctrns., 214-661-5010
		TX	<b>Houston</b> Arrow Elctrns., 713-530-4700
		TX	<b>Houston</b> Kierulff Electronics, 713-530-7030
		TX	<b>Houston</b> Schweber Elctrns., 713-784-3600
		UT	<b>Salt Lake City</b> Bell Inds. Century Elctrns. Div., 801-972-6969
		UT	<b>Salt Lake City</b> Kierulff Elctrns., 801-973-6913
		UT	<b>Salt Lake City</b> Wyle Distribution Group, 801-974-9953
		VA	<b>Richmond</b> Arrow Electronics, 804-282-0413
		WA	<b>Bellevue</b> Wyle Distribution Grp., 206-453-8300
		WA	<b>Tukwila</b> Kierulff Elctrns., 206-575-4420
		WI	<b>Brookfield</b> Schweber Elctrns., 414-784-9020
		WI	<b>Oak Creek</b> Arrow Elctrns., 414-764-6600
		WI	<b>Waukesha</b> Kierulff Elctrns., 414-784-8160
Can		Can	<b>Brampton, Ontario</b> Zentronics, 416-451-9600
Can		Can	<b>Burnaby, British Columbia</b> R.A.E. Ind. Elect. Ltd., 604-291-8866
Can		Can	<b>Calgary, Alberta</b> Cardinal Elctrns., 403-259-6817
Can		Can	<b>Calgary, Alberta</b> Zentronics, 403-272-1021
Can		Can	<b>Downsview, Ontario</b> CESCO, 416-661-0220
Can		Can	<b>Edmonton, Alberta</b> Cardinal Elctrns., 403-483-6266
Can		Can	<b>Edmonton, Alberta</b> RAE Ind. Elec., 403-451-4001
Can		Can	<b>Montreal, Quebec</b> CESCO, 514-735-5511
Can		Can	<b>Nepean, Ontario</b> CESCO, 613-226-6903
Can		Can	<b>Nepean, Ontario</b> Zentronics Ltd., 613-226-8840
Can		Can	<b>Quebec City, Quebec</b> CESCO, 418-687-4231
Can		Can	<b>Richmond, British Columbia</b> Zentronics, 604-273-5575
Can		Can	<b>St. Laurent, Quebec</b> Zentronics, 514-735-5361
Can		Can	<b>Waterloo, Ontario</b> Zentronics Ltd., 519-884-6700
Can		Can	<b>Winnipeg, Manitoba</b> Zentronics Ltd., 204-775-8661
Int'l		Int'l	<b>Australia, Burwood-Victoria</b> R & D Electronics Pty. Ltd., TEL: 288-8232
Int'l		Int'l	<b>Australia, Crows Nest-NSW</b> R & D Electronics Pty. Ltd., TEL: 439-5488
Int'l		Int'l	<b>Austria, Vienna</b> Transistor Vertriebs GmbH, TEL: (02 22) 82 94 51
Int'l		Int'l	<b>Belgium, Brussels</b> Simac Elctrns. SPRL, TEL: 02-2192453
Int'l		Int'l	<b>Denmark, Copenhagen</b> E. V. Johannsen Elektronik A-S, TEL: 45-1-839022
Int'l		Int'l	<b>Finland, Espoo</b> Nabia Elektronikka Oy, TEL: 90-46 28 29
Int'l		Int'l	<b>France, Antony</b> C.C.I., TEL: (1) 666.21.82
Int'l		Int'l	<b>France, Savres</b> Tekelec-Airtronic, TEL: (1) 534.75.35
Int'l		Int'l	<b>Hong Kong, Kowloon</b> Conmos Prods. Ltd., TEL: 3-7560103-8
Int'l		Int'l	<b>Hong Kong, Kowloon</b> Electocon Products Ltd., TEL: 3-687214-6
Int'l		Int'l	<b>India, Bombay</b> Zenith Elctrns., TEL: 384214

LINEAR-Consumer Circuits

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Audio Circuits</b>				Noise reduction, DNR				Power Amplifier, Single, <5 Watts (Cont'd)			
Amplifier and Preamplifier, Low Power				LM832				LA4220 Sanyo			
TCA210 Signetics				National				LA4430 Sanyo			
Amplifier, Headphone				Noise Reduction, Dolby				TBA820 SGS 115			
LA4170 Sanyo				HA11226 Hitachi				TBA820M SGS			
LA4175 Sanyo				LM1111 National 50				TCA830S SGS			
LA4177 Sanyo				LM1121A National				TDA1013A Signetics			
Amplifier, Hearing Aid				LM1121B National				TDA1015 Signetics			
CS263 Cherry 5				LM1121C National				ULN-2280B Sprague			
TCA1003 ITT				LM1131A National				ULN-2283B Sprague 120			
TCA1004 ITT				LM1131B National				ULN-3705M Sprague			
MPS5003 MicroPwr				LM1131C National 55				TCA830 Telefunken			
MPS5004 MicroPwr				LM1894 National				U410 Telefunken			
MPS5053 MicroPwr				NE645 Signetics				U411 Telefunken			
OM200 Panasonic				NE646 Signetics				LM388-3 TI 125			
Amplifier, with AGC, for Recorders				NE648 Signetics				SN16923 TI			
HA1319 Hitachi				NE649 Signetics				SN76000 TI			
HA1361 Hitachi				NE650 Signetics				SN76002 TI			
TDA7137 Toshiba				NE651 Signetics				SN76011 TI			
TDA7137-ST Toshiba				NE652 Signetics				SN76021 TI 130			
Attenuator, Digital				NE653 Signetics				SN76024 TI			
AD7115K AD (2827)				NE654 Signetics				TA7140 Toshiba			
Attenuator, Digital (logarithmic D/A converter)				NE660 Signetics				TA7207 Toshiba			
AD7111K AD (2827)				Noise Reduction, Dual Dolby B-Type				TA7208 Toshiba			
AD7111L AD (2827)				NE657 Signetics				CA3094 † RCA (3598) 135			
AD7111T † AD (2827)				Pop Noise Canceller				Power Amplifier, Single, 5-10 Watts			
AD7111U † AD (2827)				AN6135 Panasonic				HA1366 Hitachi			
AD7118T † AD (2827, 2829)				Power Amplifier Driver				HA1368 Hitachi			
AD7118U † AD (2827, 2829)				ICL8063C Intersil				HA1371 Hitachi			
Attenuator, Digital (with loudness compensation switch), 0 to 88.5 dB Attenuation in 1.5 dB Steps				ICL8063M † Intersil				HA1372 Hitachi			
AD7110K AD (2827, 2829)				MC3221 Motorola				HA1389 Hitachi			
AD7118K AD (2827, 2829)				MC3320 Motorola				TBA800 Hitachi			
AD7118L AD (2827, 2829)				MC3321 Motorola				TBA810 Hitachi			
Attenuator, Dual				LM391 National				TDA2002 Hitachi			
LA2600 Sanyo				Power Amplifier Driver, Dual				TDA2002 Motorola			
LC7500 Sanyo				STK3042 Sanyo 75				TDA2002A Motorola			
Audio Noise Reduction, Compander. See Linear-Telecommunications Circuits				STK3062 Sanyo				LM2002 National			
Controls (loudness, treble, bass)				STK3082 Sanyo				LM383 National			
TDA4290-5 Siemens				Power Amplifier, Single, <5 Watts				LM383A National			
Controls, Stereo, DC Operated				TBA641 Fairchild				LM384 National			
LM1035 National				TBA820 Fairchild				TDA2002 National			
TCA730A Signetics				HA1325 Hitachi				TDA2002A National			
TCA740A Signetics				HA1329 Hitachi				TDA2003 National			
TDA1074A Signetics				MC1454 Motorola				CA2002 RCA			
TDA1524 Signetics				MC1554 † Motorola				CA3131 RCA			
Digital Audio ADC				LM1879 National				CA3132 RCA 155			
PCM75 Burr-Brown (2849)				LM2001 National				LA4140 Sanyo			
Digital Audio DAC				LM2895 National				LA4230 Sanyo			
PCM501 Burr-Brown				LM380 National				LA4250 Sanyo			
PCM52 Burr-Brown (2849)				LM386 National				LA4420 Sanyo			
PCM53 Burr-Brown (2849)				LM388 National				LA4422 Sanyo			
Digital Compact Disc				LM389 National				LA4440 Sanyo			
SAA7000 Signetics				LM390 National				STK4017 Sanyo			
SAA7010 Signetics				AN214 Panasonic				STK4019 Sanyo			
SAA7020 Signetics				AN374 Panasonic				STK4332 Sanyo			
Digital Compact Disc, Filter Circuit				AN7110 Panasonic				STK4352 Sanyo			
SAA7030 Signetics				AN7114 Panasonic				STK4362 Sanyo			
Digital Noise Source				AN7115 Panasonic				L149 SGS			
MM5437 National				AN7120 Panasonic				TBA800 SGS			
Distortion Suppressor				AN7130 Panasonic				TBA810 SGS			
MP201A Analogic				CA3020 † RCA				TCA940E SGS			
Equalizer, Switchable				LA4100 Sanyo				TDA1905 SGS			
TDA2000 Siemens				LA4101 Sanyo				TDA1908 SGS			
Muting Circuit (switching noise suppressor)				LA4102 Sanyo				TDA1910 SGS			
TA7324 Toshiba				LA4110 Sanyo				TDA2002 SGS			
				LA4112 Sanyo				TDA2002A SGS			
				LA4126 Sanyo				TDA2003 SGS			
				LA4137 Sanyo				TDA2006 SGS			
				LA4138 Sanyo				TDA1037 Siemens			
				LA4180 Sanyo				TDA2003 Siemens			
				LA4182 Sanyo				TDA1020 Signetics			
				LA4200 Sanyo				TDA2002 Sprague			
				LA4201 Sanyo				TDA2002A Sprague			

† Military Temperature Range (-55° to 125°C)

\* Typical value

Bold face indicates additional data is provided on the page noted.

LINEAR

Master Selection Guide

LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Audio Circuits (Cont'd)</b>				<b>Power Amplifier, Dual, &lt;5 Watts</b>				<b>Preamplifier (Cont'd)</b>			
<b>Power Amplifier, Single, 5-10 Watts (Cont'd)</b>				HA 1374 Hitachi				TDA2310 SGS			
TDA2003 Sprague				LM1877 National				TDA3410 SGS			
ULN-3701Z Sprague				LM1895 National				TDA3420 SGS			
ULN-3702Z Sprague				LM1896 National				TA7063 Toshiba			
ULN-3703Z Sprague				LM2877 National				TA7120 Toshiba			
ULN-3784B Sprague				LM2896 National				TA7122A Toshiba			
ULX-3704B Sprague				LM377 National				TA7122B Toshiba			
TBA800 Telefunken				LM378 National				TA7129 Toshiba			
TBA810 Telefunken				AN7145L Panasonic				TA7129A Toshiba			
TDA2002 Telefunken				AN7146M Panasonic				TA7136A Toshiba			
TDA2002A Telefunken				LA4120 Sanyo				TA7322 Toshiba			
TDA2003 Telefunken				LA4125 Sanyo				<b>CA3080 RCA (3598)</b>			
TDA2004 Telefunken				LA4125T Sanyo				<b>CA3080A † RCA (3598)</b>			
SN76005 TI				ULX-3783M Sprague				<b>Preamplifier, Dual</b>			
TA7205A Toshiba				TA7203 Toshiba				XR4739 Exar			
TA7217A Toshiba				TA7214 Toshiba				<b>XR4739M † Exar (3378)</b>			
TA7222A Toshiba				TA7215 Toshiba				UA739 Fairchild			
TA7238 Toshiba				<b>Power Amplifier, Dual, 5-10 Watts</b>				μA739 Fairchild			
				HA 1377 Hitachi				μA749 † Fairchild			
				LM2878 National				HA 12012 Hitachi			
				LM2879 National				HA 1452W Hitachi			
				LM379 National				LM1301 National			
				AN7145M Panasonic				LM1303 National			
				AN7156 Panasonic				LM1897 National			
				STK433 Sanyo				LM381 National			
				STK435 Sanyo				LM382 National			
				STK443 Sanyo				LM387 National			
				STK457 Sanyo				AN7311 Panasonic			
				STK459 Sanyo				LA3115 Sanyo			
				STK463 Sanyo				LA3122 Sanyo			
				STK465 Sanyo				LA3133 Sanyo			
				TDA2004 SGS				LA3155 Sanyo			
				TDA2005 SGS				LA3160 Sanyo			
				TDA1510 Signetics				LA3161 Sanyo			
				TDA1515 Signetics				TDA2320A SGS			
				TDA2005 Telefunken				NE542 Signetics			
				TA7227 Toshiba				TDA1522 Signetics			
				<b>Power Amplifier, Dual, &gt;10 Watts</b>				SN76130 TI			
				AN7145H Panasonic				SN76131 TI			
				AN7146H Panasonic				SN76149 TI			
				AN7156N Panasonic				TA7312 Toshiba			
				STK437 Sanyo				TA7325 Toshiba			
				STK439 Sanyo				TA7685 Toshiba			
				STK441 Sanyo				<b>Preamplifier, Quad</b>			
				<b>Preamplifier</b>				<b>XR4212 Exar (3378)</b>			
				MA106 AnalogSys				<b>Signal Delay</b>			
				ZN459 † Ferranti				<b>MN3001 Panasonic (3559)</b>			
				ZN459C Ferranti				<b>MN3010 Panasonic (3559)</b>			
				ZN459CP Ferranti				MN3204 Panasonic			
				ZN460 † Ferranti				MN3207 Panasonic			
				ZN460C Ferranti				<b>Sound Generator</b>			
				ZN460CP Ferranti				AY3-8910 GI			
				HA12017 Hitachi				AY3-8912 GI			
				HA1406 Hitachi				<b>Sound Generator Controller for Microprocessor-Based Systems</b>			
				HA1457 Hitachi				SN76489 TI			
				LM1837 National				SN76489A TI			
				LM3080 National				SN76493 TI			
				LM3080A National				SN76494 TI			
				μPC1023 NEC				SN76495 TI			
				μPC1024 NEC				<b>Sound Generator w/Audio Amp</b>			
				μPC1032 NEC				SN76487 TI			
				μPC566 NEC				SN76488 TI			
				μPC592 NEC				<b>Voltage Controlled Oscillator for Electronic Music</b>			
				AN360 Panasonic				SSM2038 SSM			
				AN370 Panasonic				<b>Dual DC Operated Tone/Volume/Balance Circuit</b>			
				SL561 Plessey				LM1036 National			
				SL561B Plessey				<b>Dual DC Operated Tone/Volume/Balance Circuit with Stereo Enhancement</b>			
				SL561C Plessey				LM1040 National			
				LA3110 Sanyo							
				LA3120 Sanyo							
				LA3130 Sanyo							
				LA3150 Sanyo							
				LA3170 Sanyo							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR—Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line		
<b>Automotive Circuits</b>				<b>Exposure Control</b> (Cont'd)				<b>Alarm Clock Radio</b> (Cont'd)					
Automotive Lamp Monitor	ULN-2435A	Sprague	5	CS483	Cherry	55	MM5465	National	(4118)	115			
	ULN-2445A	Sprague		CS484	Cherry		MM5466	National					
	ULN-2455A	Sprague		CS485	Cherry		<b>MSM5524</b>	<b>OKI</b>					
Clock	PCF1171	Signetics		CS486	Cherry		LM8363	Sanyo					
	PCF1172	Signetics		CS487	Cherry		LM8364	Sanyo					
Clock with Vacuum Fluorescent Display Drivers	S4003	AMI	CS488	Cherry	<b>Automotive Clock with LCD Display Drivers</b>								
	SCL5233	SSS	CS489	Cherry	PCF1171		Signetics	60					
			CS580	Cherry	PCF1172		Signetics						
High Energy Ignition Circuit	MC3334	Motorola	S0258A	Siemens	<b>Clock Automotive, with Vacuum Fluorescent Display Drivers</b>								
Ignition Circuit	MC3333	Motorola	S0289	Siemens	S2709		AMI	120					
Injector Drive Controller	LM1949	National	<b>Strobe Light Controller</b>				<b>MSM5557</b>				<b>OKI</b>		
Lamp Monitor	ULN-2401A	Sprague	CS139	Cherry	<b>Clock Circuits (See Also: Digital—CMOS Oscillators/Dividers. Linear—Other Linear Devices—Oscillator)</b>								
Low Coolant Detector	ULN-2429A	Sprague	<b>CB Radio Circuits</b>				<b>S4003</b>	<b>AMI</b>	(705)	125			
Solenoid Driver	MC3484V2	Motorola	<b>Channel Selector/Display Driver</b>				HD44010	Hitachi					
	MC3484V4	Motorola	FCB8010				ICM7045	Intersil					
Speed Control Processor	MC14460	Motorola	MN6080				ICM7051A	Intersil					
Speedometer and Mileage Indicator	SAY115	ITT	LC7181				ICM7051B	Intersil					
Traffic Information Broadcast Decoders	UAA1009	ITT	LC7184				ICM7052	Intersil					
	S0280	Siemens	LC7191				SAJ300	ITT					
	S0281	Siemens	Programmer, for PLL				C1200	LSIComp					
	S551	Siemens	MP7156				MM5309	National					
	S552	Siemens	LC7120				MM5311	National					
Voltage Regulator	TCA700	ITT	Synthesizer Set, 40 Channel, BCD Channel Setting				MM53110	National					
	MC3325	Motorola	MP7189				MM5312	National					
20-Watt Automotive Power Amplifier	LM2005	National	Synthesizer Set, 40 Channel, BCD Channel Setting				MM5313	National					
			LC7131				MM5314	National					
			LC7135				MM5315	National					
			Transceiver Receiver				MM5318	National					
			TMS1022				TI						
			<b>Clock/Watch Circuits</b>				MM5378	National					
			Alarm Clock				<b>MSM5509</b>	<b>OKI</b>					
			HD38980				Hitachi	(4118)					
			HD38991C				Hitachi						
			HD43115				Hitachi	(4118)					
			HD43890				Hitachi						
			ICM1115				Intersil	(4118)					
			ICM1115A				Intersil						
			ICM1115B				Intersil	145					
			ICM7038A				Intersil						
			ICM7038B				Intersil	150					
			ICM7049A				Intersil						
			ICM7050				Intersil	155					
			MM53108				National						
			MM5316				National	160					
			MM5387				National						
			MM5402				National	165					
			MM5405				National						
			MM58143				National	170					
			MM58144				National						
			μPD5388				NEC	175					
			μPD833				NEC						
			MN6076				Panasonic	180					
			MN6092				Panasonic						
			LM8360				Sanyo	185					
			LM8361				Sanyo						
			MB561				Signetics	190					
			MB562				Signetics						
			<b>Alarm Clock Radio</b>				<b>Color Bar Generator</b>						
			ICM7223				Intersil	MM5322	National				
			ICM7223A				Intersil	<b>Color Control for VIR Broadcasts</b>					
			ICM7223VF				Intersil	HA11409	Hitachi				
			MM53113				National	<b>Digital Clock, Channel and Time Display</b>					
			MM53124				National	MM58106	National				
			MM5455				National	<b>Digital Tuner</b>					
			MM5456				National	MM58142	National				
			MM5457				National	<b>Melody Circuit, for Clocks</b>					
			MM5462				National	7910	SMOS				
			MM5463				National	7920	SMOS				
			MM5468				National	7930	SMOS				
			MM5464				National	SVM7940	SMOS				
								SVM7950	SMOS				
								SVM7960	SMOS				
								SVM7970	SMOS				
								SVM7990	SMOS				

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR

Master Selection Guide

LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line	
<b>Clock/Watch Circuits (Cont'd)</b>				<b>Organ Circuits</b>								
Melody Circuit, Westminster Chime				Delay Circuit, See also Analog Shift Registers and Serial Analog Delay under Linear-Other Devices.				Mixer, IF, AGC				
	SVM5530	SMOS			TDA1022	Signetics			ULN-2243A	Sprague	105	
	SVM7810	SMOS						Radio System for Electronically Tuned Radios				
Stopwatch/Clock (LCD)				Digital Accompaniment Interface				Receiver System†				
	ICM7045	Intersil			M109	SGS			HA1151	Hitachi		
	WD2412	Western		Electronic Attenuator						HA1197	Hitachi	
Stopwatch (LED)				Frequency Divider, 6-Stage						HA1199	Hitachi	
	ICM7205	Intersil	5		S10131	AMI	60		μPC1215V	NEC	110	
	ICM7215	Intersil		Frequency Divider, 7-Stage						CA3088	† RCA	
Watch, Analog (See also Digital—CMOS Oscillators/Dividers)				Frequency Divider, 8-Stage, with Staircase Generator						CA3123	† RCA	
	HC43850	Hitachi			MN6004	Panasonic			TCA440	Siemens		
	ICM7047	Intersil		Frequency Generator. See also Organ, Top Octave Generator						TDA1046	Siemens	
	ICM7245	Intersil			MN6005	Panasonic	70		TDA1072	Signetics	115	
Watch (LCD), Digital				Key Divider						TEA5550	Signetics	
	HC42022	Hitachi	10		S10430	AMI			ULN-2249A	Sprague		
	HC42030	Hitachi		Monophonic Synthesizer						ULN-3838A	Sprague	
	HC43032	Hitachi			M110	SGS	65		TDA1072	Telefunken		
	HC43802	Hitachi		Noise Generator						SN76635	TI	120
	HC43803	Hitachi			S2688	AMI			TA7306	Toshiba		
	ICM1424B	Intersil			MMS837	National			TA7615	Toshiba		
	ICM1424MB	Intersil		Rhythm Circuit						LM3820	National	
	ICM1424MC	Intersil			S2567	AMI			ZN414	Ferranti		
	ICM7210	Intersil			LM8372	Sanyo	15		SL623C	Plessey	125	
	ICM7210C	Intersil			LM8471	Sanyo			LM1981	National		
	ICM7210M	Intersil			LM8972	Sanyo			Stereo Demodulator			
	ICM7210MC	Intersil			M258	SGS			HS3604	Harris		
	ICM7220	Intersil			M259	SGS			Tuner for Car Radio			
	ICM7220B	Intersil			M268	SGS			TA7402	Toshiba		
	ICM7220M	Intersil			M269	SGS			TA7616	Toshiba		
	ICM7220MA	Intersil		Solo and Accompaniment						Tuner System		
	ICM7220MC	Intersil			M108	SGS			LA1130	Sanyo	130	
	ICM7270	Intersil		Tone Generator						LA1240	Sanyo	
	ICM7272	Intersil			M082	SGS			LA1245	Sanyo		
	<b>MSM5803</b>	<b>OKI</b>	<b>(4118)</b>		M082AA	SGS			<b>Radio Circuits, AM/FM</b>			
	MS680	RTC			M083	SGS			AM/FM Phase and Signal Locked Loop (PSL <sup>2</sup> ) Tuner			
	MS681	RTC			M083AA	SGS			LC7210			
	MS682	RTC			M086	SGS			AM/FM PLL Synthesizer Tuner			
	MS683	RTC			M086AA	SGS			LC7220			
	LC5621	Sanyo		Top Octave Generator						LC7225		
	PCA1122	Signetics			S50240	AMI			AM/FM/SW Frequency Counter			
	SCL54301	SSS			S50241	AMI			MSM5526			
	SCL5482	SSS			S50242	AMI			Display Driver for Receiving Frequency Digital Display			
	SCL5483	SSS			MK50240	Mostek			TD6301A			
Watch (LCD), Digital Chronograph				Toy Organ						FM/SW/MW/LW Frequency Display		
	ICM7220C	Intersil	40		SAA1900	ITT			LC7253			
	MP7160	MicroPwr		Dual Voltage Controlled Amplifier						LC7257		
	LC5613	Sanyo			SSM2020	SSM			LC7258			
Watch (LCD), Digital Chronograph/Alarm				Radio Circuits, AM						Frequency Counter		
	H17000	Holt			SSM2022	SSM			<b>MSM5525</b>			
	MP7216	MicroPwr		AM/SW Tuner						OKI (4118)		
Watch (LCD), Digital with Alarm				Amplifier with Demodulator and Volume Control						IF Amplifier		
	H17020	Holt	45		TDA1048	Siemens			μA757			
	ICM7220A	Intersil		CQUAM AM Stereo Decoder						LM1821		
	ICM7220FA	Intersil			MC13020	Motorola			LM1868			
	ICM7220MFA	Intersil		Frequency Synthesizer						National		
	ICM7221	Intersil			TC9131	Toshiba			National			
	ICM7222	Intersil	50	Mixer, IF, AGC						Fairchild		
	MP7150	MicroPwr		Mixer, IF, AGC						National		
Watch (LCD), Digital with Alarm and Music Enable				Mixer, IF, AGC						National		
	ICM7271	Intersil		Mixer, IF, AGC						NEC		
	ICM7273	Intersil		Mixer, IF, AGC						Panasonic		
Watch (LED), Digital				Mixer, IF, AGC						AN7266		
	ICM7214	Intersil		Mixer, IF, AGC						AN7266		
	ICM7214A	Intersil	55	Mixer, IF, AGC						LA1201		
				Mixer, IF, AGC						TAA991		
				Mixer, IF, AGC						Radio Frequency Display		
				Mixer, IF, AGC						MM5430		
				Mixer, IF, AGC						MM5431		
				Mixer, IF, AGC						Radio Frequency Display Driver and Clock		
				Mixer, IF, AGC						HA12404		
				Mixer, IF, AGC						HA12405		
				Mixer, IF, AGC						T1400		
				Mixer, IF, AGC						Receiver System		
				Mixer, IF, AGC						HA11211		
				Mixer, IF, AGC						HA11251		
				Mixer, IF, AGC						HA12402		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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(Continued)



LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Radio Circuits, AM/FM (Cont'd)</b>				<b>IF Amplifier/Detector (Cont'd)</b>				<b>Receiver System with Squelch and Scan Control</b>			
Receiver System				LM3011	National			MP5071	MicroPwr		120
	LM1866	National		LM3075	National			MC3357	Motorola		
	AN217P	Panasonic		LM3089	National			RF/IF Amplifier	CA3127	RCA	
	AN253	Panasonic		LM3189	National			TBA120U	Signetics		
	AN366	Panasonic		TBA120	National		65	TCA770	Signetics		
	LA1210	Sanyo	5	μPC1028	NEC			Stereo Decoder	XR1310	Exar	125
	TDA1220A	SGS		TBA120	Plessey			HA11223	Hitachi		
	TEA5570	Signetics		TBA750	Plessey			HA11227	Hitachi		
	TDA1083	Sprague		CA2111A	† RCA		70	HA1156	Hitachi		
	TDA1090	Sprague		CA3013	RCA			HA1196	Hitachi		
	ULN-2204A	Sprague	10	CA3014	† RCA			MC1309	Motorola		130
	ULN-2240A	Sprague		CA3042	RCA			MC1310	Motorola		
	ULN-2241A	Sprague		CA3065	RCA			TCA4500	Motorola		
	ULN-2242A	Sprague		CA3075	† RCA		75	μA758A	Motorola		
	ULX-3840A	Sprague		CA3089	† RCA			LM1310	National		
	TDA1083	Telefunken	15	CA3189	RCA			LM1800	National		135
	U416	Telefunken		CA3209	RCA			LM1870	National		
	U417	Telefunken		CA3215	RCA			LM4500A	National		
	U418	Telefunken		LA1140	Sanyo			μPC1026	NEC		
	TA7613	Toshiba		LA1150	Sanyo		80	AN362	Panasonic		140
	TA7614	Toshiba	20	LA1230	Sanyo			CA1310A	RCA		
Search Tuning	TDA1580	Signetics		LA1231N	Sanyo			CA3195	RCA		
Signal Processor				TDA1200	SGS			CA758	RCA		
	ULX-3804A	Sprague		S041	Siemens			LA3301	Sanyo		
Synthesizer (FM prescaler, preamplifier and reference oscillator)	AC5945	TI		TBA120S	Siemens		85	LA3350	Sanyo		
Tuner Control	LC7200	Sanyo		TDA1047	Siemens			LA3361	Sanyo		145
	LC7201	Sanyo		CA3089	Signetics			LA3365	Sanyo		
	LC7203	Sanyo	25	CA3089D2	Signetics			LA3370	Sanyo		
	LC7207	Sanyo		CA3189	Signetics			LA3375	Sanyo		
Tuner, Radio Frequency Display and Clock								LA3380	Sanyo		
	LC7250	Sanyo		IF Amplifier Detector			90	LA3381	Sanyo		150
TV Synthesizer	SAA1057	Signetics		TBA120U	Signetics			LA3390	Sanyo		
<b>Radio Circuits, FM</b>				TBA750C	Signetics			TCA4500A	Siemens		
Detector. See Linear-Phase Locked Loop Circuits and IF Amplifier/Detector below.			30	TCA770	Signetics			TCA4510	Siemens		
Front End	AN7213	Panasonic		ULN-2111A	Sprague			LM1870	Signetics		
	TDA1571	Signetics		ULN-2136A	Sprague			TDA1578A	Signetics		155
	TDA1574	Signetics		ULN-3889A	Sprague		95	TEA5580	Signetics		
Front End Timer				TBA120	Telefunken			μA758	Signetics		
	TDA1094	Telefunken		TDA1093	Telefunken			ULN-3809A	Sprague		
	TA7335	Toshiba	35	SN76642	TI			ULN-3810A	Sprague		160
IF Amplifier	HA1201	Hitachi		SN76643	TI		100	ULN-3812A	Sprague		
	HA1211	Hitachi		SN76669	TI			SN76104	TI		
	MC1350	Motorola		SN76675	TI			SN76105	TI		
	AN278	Panasonic		SN76689	TI			SN76110	TI		
	AN377	Panasonic						SN76111	TI		165
	CA3002	† RCA		IF Amplifier, Narrow Band			105	SN76113	TI		
	CA3005	† RCA (3603)		MC3359	Motorola			SN76115	TI		
	CA3006	† RCA (3603)		ULN-3859A	Sprague			SN76116	TI		
	CA3012	† RCA		IF Amplifier, PLL, Squelch, Narrow Band				TA7157A	Toshiba		
	CA3028	† RCA (3603)	45	SL6640	Plessey			TA7401A	Toshiba		
	CA3076	† RCA		IF System				TA7604A	Toshiba		170
	LA1111	Sanyo		LA1235	Sanyo			TA7624	Toshiba		
	LA1222	Sanyo		LA2300	Sanyo			Stereo Multiplexer (PLL)			
	TDA4200	Siemens						TA7343	Toshiba		
	TDA1576	Signetics		Interference Absorption Circuit				Tuner			
	TEA5560	Signetics		TDA1001B	Signetics			TUA1000	Siemens		
	ULN-2243A	Sprague		LED Tuning Indicator				TDA1062	Telefunken		
	SN76676	TI		LB1450	Sanyo			TDA1062S	Telefunken		175
	SN76678	TI		LB1473	Sanyo						
	TA7302	Toshiba		Mixer, IF, AGC System				<b>Remote Control Circuits</b>			
	TA7303	Toshiba	55	LA1160	Sanyo			Amplifier-Detector			
IF Amplifier/Detector				Modulator Circuit (VCO/modulator, 1.4 to 14 MHz)				MC3373	Motorola		
	HA11225	Hitachi		MC1376	Motorola			Receiver, Infrared			
	HA1137	Hitachi		Noise Canceller				TDA3047	Signetics		
	HA12411	Hitachi		HA11219	Hitachi			Receiver Infrared			
	LM1865	National	60	LA2100	Sanyo			TDA3048	Signetics		
				LA2101	Sanyo			Remote Control, Infrared			
				LA2110	Sanyo			SAA3004	Signetics		
				Prescaler (for digital tuning)				SAA3027	Signetics		180
				TD6104	Toshiba						
				Receiver System							
				TDA7000	Signetics (3663)						
				TDA7010T	Signetics (3666)						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line	
<b>Remote Control Circuits (Cont'd)</b>				<b>Transmitter/Receiver (encoder/decoder), 2 device sets (Cont'd)</b>				<b>Tape Recorder Circuits, Audio</b>				
RF Amplifier/Mixer				65				Amplifier System				
	CA3005	† RCA		SN76730	TI			TA7223	Toshiba			
	CA3006	† RCA		SN76831	† TI			TA7628	Toshiba			
	CA3049	† RCA		<b>SN76832A</b>	<b>TI</b>	<b>(387)</b>		Amplifier System (for cassette recorders)			115	
Transceiver (ultrasonic)				70				TA7628H				Toshiba
	LM1812	National		TC9132	† Toshiba		Amplifier, with AGC, for Recorders					
Transmitter, Infrared				75				TDA2054				SGS
	MM53226	National		TC9133	Toshiba		TA7137				Toshiba	
	SAA3006	Signetics		TC9134	Toshiba		TA7137-ST				Toshiba	
	<b>SN76881</b>	<b>TI</b>	<b>(387)</b>	<b>Speech Circuits</b>				Audio System				120
	<b>SN76882</b>	<b>TI</b>	<b>(387)</b>	Delta Modulation System, Continuously Variable Slope				HA12006				Hitachi
	<b>SN76891</b>	<b>TI</b>	<b>(387)</b>	HC55536				LM1818				National
NOTE: In the following category ONLY, the dagger symbol in front of a company name indicates a transmitter/encoder. Absence of a dagger designates a receiver/decoder.				HC55564				AN262				Panasonic
Transmitter/Receiver (encoder/decoder), 2 device sets				10				Automatic Cuing Circuit				
	S2600	† AMI		Preprogrammed Vocabulary ROM				LC7510				Sanyo
	S2601	AMI		VM61001				LC7512				Sanyo
	S2602	† AMI		<b>VM61002</b>				Automatic Leveling				
	S2603	AMI		<b>VM61003</b>				S028-2				Siemens
	S2604	† AMI		<b>VM61004</b>				Automatic Program Search				
	S2605	AMI		<b>VM61005</b>				TC9138				Toshiba
	S2742	AMI		VM71001				TC9139				Toshiba
	S2743	† AMI		VM71002				Automatic Reverse/Eject for car stereo				
	S2747	AMI		<b>VM71003</b>				TD6304				Toshiba
	S2748	AMI		Speech Analyzer/Synthesizer, ADPCM				TD6308				Toshiba
	AY3-8470	† GI		<b>MSM5218</b>				Auto-Reverse Cassette Control				
	AY3-8475	GI		TCM1705A				TDA1508				Signetics
	SAA1250	† ITT		Speech Circuit for Electronic Telephones				Broadband Compressor				
	SAA1251	ITT		DT1051				U401				Telefunken
	SAA1272	ITT		DT1054				Controller (mode change, absence recording)				
	SAA1350	† ITT		Speech FIFO Buffer Memory and Control Logic				HA12001				Hitachi
	SAA1351	ITT		SPB640				TC9121				Toshiba
	MC14457	† Motorola		Speech Generator				Head, Microphone, Headphone, Buffer, and Record Amplifiers				125
	MC14458	Motorola		UAA1003				TA7617				Toshiba
	MC145026	Motorola		UAA1103				IC Switch (cassette)				
	MC145027	Motorola		Speech Interface Control Logic				TC9143				Toshiba
	MC145028	Motorola		SPR000				TC9144				Toshiba
	LM1871	National		Speech Synthesis Processor				Motor Speed Control (automatic stop, dc manual stop, biasing and erasing oscillator)				
	LM1872	National		SP-0250				TDA7770				SGS
	MM5425	National		SP0256				Motor Speed Control, Stop, Pause, Cassette Ejection				
	MM58250	† National		DT1000				TDA7270				SGS
	<b>MSL9362</b>	<b>OKI</b>	<b>(4118)</b>	MM54104				Power Amp, Preamp with ALC, Meter Driver				
	<b>MSL9363</b>	<b>OKI</b>	<b>(4118)</b>	SVM9300				TA7225				Toshiba
	MN6021	Panasonic		<b>ME8000</b>				Power Amp, Preamp with ALC, Meter Driver, Mute Control				
	MN6025	Panasonic		TMS5100				TA7224				Toshiba
	ML920	Plessey		TMS5200				Power Amplifier (cassette)				
	ML922	Plessey		SC-01				TA7229				Toshiba
	ML926	Plessey		Speech Synthesis System, Basic Numbers Kit				Power Amplifier (minicassette)				
	ML928	Plessey		DT1052				TA7331				Toshiba
	ML929	Plessey		DT1055				Preamp with ALC				
	SL490	Plessey		DT1050				LA3201				Sanyo
	M1124	† SGS		DT1053				LA3210				Sanyo
	SAB3209	Siemens		DT1056				SSM2011				SSM
	SAB3210	† Siemens		DT1057				Preamp with ALC (cassette)				
	SAB3271	Siemens		Speech Synthesis System, Standard Vocabulary Kit				TDA1054M				SGS
	SAB4209	Siemens		DT1050				Preamp with ALC (minicassette)				
	SDA2008	† Siemens		DT1053				TA7330				Toshiba
	SDA3205	Siemens		DT1056				System (power amp, preamp with ALC, for minicassette recorders)				
	SDA3206	† Siemens		DT1057				TA7625				Toshiba
	SAA5000	† Signetics		Speech Synthesizer, ADPCM				Tape Counter				140
	SAA5012	Signetics		<b>MSM5205</b>				LM8523				Sanyo
	SAF1032	Signetics		<b>MSM6202</b>				Tape Deck Amplifier System				
	SAF1039	Signetics		Speech Synthesizer, Linear Predictive Data Stored in ROM				TA7663				Toshiba
	U318	Telefunken		<b>S3610</b>				Tape Deck Amplifier, Dual Channel				
	U321	† Telefunken		<b>S3620</b>				TA7639				Toshiba
	U327	† Telefunken		Synthesized-Speech ROM				Tape Deck Automatic Reverse				
	U328	Telefunken		<b>TMS6100</b>				AN6249				Panasonic
	SN76605	† TI		<b>TMS6125</b>				AN6250				Panasonic
	SN76606	TI										

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Tape Recorder Circuits, Audio (Cont'd)</b>				<b>Camera Sync. Generator (Cont'd)</b>				<b>Chroma Circuits (demodulators, IF amplifiers, luminance control, signal processors, and various combinations) (Cont'd)</b>			
Tape Deck Key Controller	AN6251	Panasonic		HD44007	Hitachi			HA 11580	Hitachi		
<b>Tape Recorder Circuits, Video</b>				MM5321	National		55	MC1377	Motorola		
Color Processing Circuit				MN6063	Panasonic			MC1399	Motorola		
HA11704	Hitachi			MN6064	Panasonic			LM1828	National		120
HA11705	Hitachi			S178A	Siemens			LM1848	National		
HA11706	Hitachi			<b>CATV Amplifiers (hybrid)</b>				LM2887	National		
AN305	Panasonic		5	BGY54A	Amperex			LM3126	National		
AN337	Panasonic			BGY55A	Amperex		60	TBA510	National		125
MN6061A	Panasonic			BGY56	Amperex			TBA530	National		
<b>Dropout Compensation Circuit</b>				BGY57	Amperex			TBA540	National		
AN316	Panasonic			BGY58A	Amperex			TBA560	National		
<b>FM Limiter</b>				BGY74	Amperex			TBA990	National		
HA11703	Hitachi		10	BGY75	Amperex			TDA2522	National		
AN304	Panasonic			BGY78	Amperex			TDA2523	National		130
<b>FM Recording Signal Processor</b>				MHW1121	Motorola		65	TDA2530	National		
HA11701	Hitachi			MHW1127	Motorola			TDA2560	National		
AN6321	Panasonic			MHW1171	Motorola			TDA3500	National		
AN6331	Panasonic			MHW1172	Motorola			TDA3501	National		
<b>Head Amplifier</b>				MHW1182	Motorola		70	UPC580	NEC		135
AN6320	Panasonic			MHW1222	Motorola			μPC1380	NEC		
AN6330	Panasonic		15	MHW1341	Motorola			AN236	Panasonic		
<b>Luminance Processing</b>				MHW1342	Motorola			AN5310	Panasonic		
AN302	Panasonic			MHW1343	Motorola			AN5311	Panasonic		140
AN303	Panasonic			MHW1391	Motorola		75	AN5320	Panasonic		
<b>Playback Signal Processor</b>				MHW1392	Motorola			AN5330	Panasonic		
HA11702	Hitachi			MHW5181	Motorola			AN5620	Panasonic		
<b>Servo Signal Processing</b>				MHW5182	Motorola			AN5630	Panasonic		
HA11707	Hitachi			MHW559	Motorola		80	TDA2532	Plessey		
AN301	Panasonic		20	MHW560	Motorola			CA1398	RCA		145
AN318	Panasonic			MHW561	Motorola			CA3067	RCA		
AN6341	Panasonic			MHW562	Motorola			CA3070	RCA		
AN6342	Panasonic			MHW570	Motorola			CA3071	RCA		
AN6344	Panasonic			MHW572	Motorola		85	CA3072	RCA		
<b>Video Recording Processing</b>				MHW580	Motorola			CA3121	RCA		150
AN6300	Panasonic		25	MHW594	Motorola			CA3125	RCA		
<b>16-Channel Precision Timer/Driver</b>				CA2100	TRW-LSI			CA3126	RCA		
CD22401	RCA	(838)		CA2300	TRW-LSI		90	CA3128	RCA		
<b>Television Circuits</b>				CA2301	TRW-LSI			CA3135	RCA		
AFT System	HA1108	Hitachi		CA2418	TRW-LSI			CA3137	RCA		155
HA1126	Hitachi			CA2600	TRW-LSI			CA3143	RCA		
MC6190	Motorola		30	CA2601	TRW-LSI			CA3144	RCA		
MC6191	Motorola			<b>Channel Display Driver/Decoder</b>				CA3145	RCA		
MC6192	Motorola			LM1017	National		95	CA3151	RCA		
MC6193	Motorola			U143	Telefunken			CA3153	RCA		160
MC6194	Motorola			<b>Channel Selection</b>				CA3156	RCA		
MC6195	Motorola			μPC1009	NEC			CA3158	RCA		
MC6196	Motorola		35	AN5010	Panasonic			CA3170	RCA		
LM3064	National			AN5011	Panasonic			CA3172	RCA		
AN321	Panasonic			LB1500	Sanyo		100	CA3194	RCA		165
CA3064	† RCA			LB1515	Sanyo			CA3201	RCA		
CA3139	RCA			LB1550	Sanyo			CA3217	RCA		
LA1364	Sanyo		40	SN76701	TI			CA3221	RCA		
SDA4260	Siemens			SN76702	TI			LA1368	Sanyo		170
SN76565	TI			SN76710	TI			LA1369	Sanyo		
TA7070	Toshiba			SN76711	TI		105	LA1390	Sanyo		
<b>Bandpass Filter</b>				SN76721	TI			LB1332	Sanyo		
2002Z	NCM			SN76727	TI			TDA2140	SGS		
<b>Cable TV Channel Processor</b>				TA7171	Toshiba			TDA2151	SGS		
SDA2110	Siemens		45	TA7172	Toshiba			TDA2161	SGS		175
<b>Camera</b>				TA7177	Toshiba		110	TDA2522	Siemens		
AN614	Panasonic			TA7178	Toshiba			TDA2530	Siemens		
AN616	Panasonic			<b>Character Generator (indicates tuning voltage scale and band info on screen)</b>				TDA2560	Siemens		
<b>Camera Sync. Generator</b>				U191	Telefunken			TDA3500	Signetics		
3262A	Fairchild			<b>Chroma Circuits (demodulators, IF amplifiers, luminance control, signal processors, and various combinations)</b>				TDA3505	Signetics		180
3262B	Fairchild			HA11247	Hitachi			TDA3510	Signetics		
ZNA134E	Ferranti		50	HA11409	Hitachi		115	TDA3520	Signetics		
ZNA134J	† Ferranti			HA11412	Hitachi			TDA3562	Signetics		
<b>(Continued)</b>				HA11417	Hitachi			TDA3563	Signetics		
<b>(Continued)</b>				<b>(Continued)</b>				TDA4550	Signetics		185
<b>(Continued)</b>				<b>(Continued)</b>				TDA4560	Signetics		

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\* Typical Value

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LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Television Circuits (Cont'd)</b>											
Chroma Circuits (demodulators, IF amplifiers, luminance control, signal processors, and various combinations) (Cont'd)											
	ULN-2260A	Sprague		Frequency Synthesis Control (interface between synthesizer and varicap diode tuner)	MC2801	Motorola		Modulator Circuit for Video Tape Recorders and Disc Players	MC1374	Motorola	
	ULN-3914A	Sprague		Frequency Synthesizer	SAB3035	Signetics	60	Noise Reduction System	TA7652	Toshiba	
	TBA520	Telefunken		Frequency Synthesizer, interfaces to MPU	MN6044	Panasonic		On Screen Channel Display	M191	SGS	125
	TBA530	Telefunken			MN6049	Panasonic			M192	SGS	
	TBA540	Telefunken	5	Games, See Consumer Circuits, Video Games				On Screen Display	SDA2005	Siemens	
	TBA560	Telefunken		Horizontal & Vertical Processor				On Screen Tuning Voltage Bargraph Display	UAA190	Siemens	
	TBA990	Telefunken		HA11235	Hitachi			PAL Color Encoder and Video Summer	TEA1002	Signetics	
	TDA2140	Telefunken		HA11244	Hitachi			Pattern Generator	4001Z	NCM	130
	TDA2150	Telefunken		LM1880	National			Picture Contrast/Brightness	TA7126	Toshiba	
	TDA2151	Telefunken	10	AN295	Panasonic				TA7134	Toshiba	
	TDA2160	Telefunken		AN5410	Panasonic			Power Supply Controller	LA5112	Sanyo	75
	TDA2161	Telefunken		AN5411	Panasonic			Prescalers			135
	SN76242	TI		AN5430	Panasonic			Program Memory (17x16)	M193	SGS	
	SN76243	TI		AN5431	Panasonic			Program Memory (20x12), D/A	SN76720	TI	
	SN76246	TI	15	AN5435	Panasonic			Radio	M755	SGS	
	SN76267	TI		CA3202	RCA				M756	SGS	
	SN76298	TI		CA3223	RCA			Signal Processor, Bilingual	CA3120	RCA	140
	SN76645	TI		TDA2577A	Signetics				CA3142	RCA	
	SN76889	TI		TDA2578A	Signetics				CA3154	RCA	
	TA7141A	Toshiba	20	TDA2595	Signetics				LA7751	Sanyo	
	TA7148	Toshiba		Horizontal Processor					LA7755	Sanyo	
	TA7149	Toshiba		TBA940	ITT				TBA890	Signetics	145
	TA7150	Toshiba		TBA950	ITT				SN76524	TI	
	TA7161	Toshiba		TDA1950	ITT				SN76544	TI	
	TA7167	Toshiba		TDA9500	ITT				SN76545	TI	
	TA7168	Toshiba	25	TDA9503	ITT			Small Signal Combination IC for Monochrome TV	TDA4500	Signetics	150
	TA7169	Toshiba		TDA9513	ITT			Sound Circuits	HA11229	Hitachi	
	TA7173	Toshiba		MC1391	Motorola				HA1124	Hitachi	
	TA7174	Toshiba		MC1394	Motorola				HA1125	Hitachi	
	TA7192	Toshiba	30	LM1391	National				HA1364	Hitachi	
	TA7193	Toshiba		TBA920	National				TDA1235	ITT	
	TA7607A	Toshiba		TBA950-2	National				TDA1236	ITT	
	TA7611A	Toshiba		TDA2590	National				MC1357	Motorola	
	TA7621	Toshiba		TDA2591	National				MC1358	Motorola	
	TA7622A	Toshiba	35	AN5750	Panasonic				TDA1190	Motorola	
	TA7650	Toshiba		AN5760	Panasonic				TDA1190P	Motorola	160
	TA7659	Toshiba		TDA2591	Plessey				TDA3190P	Motorola	
	TA7660	Toshiba		CA1391	RCA				TBA120S	National	
	TA7661	Toshiba		CA1394	RCA				μPC575	NEC	
	TA7675	Toshiba	40	CA3159	RCA				AN240P	Panasonic	
	TA7676	Toshiba		CA3210	RCA				AN241	Panasonic	
	TA7678	Toshiba		CA920A	RCA				AN340	Panasonic	165
				TDA1180	SGS				AN5210	Panasonic	
D/A Converter (converts an input to coarse and fine tune outputs)	SAB3013	Signetics		TDA2593	Signetics				AN5220	Panasonic	
Decoder, 1 of 16	M054	SGS		TDA2594	Signetics				AN5250	Panasonic	
	M055	SGS	45	TDA2591	Telefunken				AN5260	Panasonic	
Deflection Correction Circuit	TDA4610	Siemens		TDA2593	Telefunken				AN5730	Panasonic	
Digital Tuner and Signal Detector	LM1019	National		SN76525	TI				TBA120	Plessey	
	TDA4431	SGS		TA7151	Toshiba		105		TBA750	Plessey	
Digital Tuning	SAB3035	Signetics		Intercarrier Frequency Modulator	4002Z	NCM			CA1190	RCA	
	SAB3036	Signetics		Line Selector	4003Z	NCM			CA1191	RCA	175
	SAB3037	Signetics	50	MATV Amplifiers (hybrid)					CA3042	RCA	
Digital Tuning Device Set	SAA1075	ITT		OM320	Amperex				CA3065	RCA	
	SAA1274	ITT		OM321	Amperex				CA3075	† RCA	
	SAA1276	ITT		OM322	Amperex		110		CA3134	RCA	
Digital Tuning System Memory Controller	TA7619A	Toshiba	55	OM323	Amperex				LA1320	Sanyo	180
Digital Tuning System (PLL)	TC9136	Toshiba		OM323A	Amperex						
	TC9137	Toshiba		OM335	Amperex						
Frequency Synthesis Control (converts tuning data into frequency information for SDA100 system)	SM564	Siemens		OM336	Amperex						
				OM337	Amperex		115				
				OM337A	Amperex						
				OM339	Amperex						
				OM345	Amperex						
				OM350	Amperex		120				
				OM360	Amperex						
				OM361	Amperex						
				OM370	Amperex						

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LINEAR-Consumer Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line	
<b>Television Circuits (Cont'd)</b>				<b>Tuner Controller, Non-Volatile Storage for 16 Programs U193</b>				<b>Video Circuits (IF amplifiers, detectors, AGC, AFC, and various combinations) (Cont'd)</b>				
<b>Sound Circuits (Cont'd)</b>				<b>Tuner Controller (2 device set)</b>								
	LA1363	Sanyo	5		SAA1020	ITT	60		LM2889	National	120	
	LA1365	Sanyo				SAA1121		ITT		TBA440		National
	TDA1190	SGS				SAA1220		ITT		TBA970		National
	TDA2190	SGS								TDA2540		National
	TDA3190	SGS			<b>Tuning System</b>	SDA100		Siemens		TDA2541		National
	TDA2840	Siemens				SDA200		Siemens		TDA440		National
	TDA2841	Siemens			<b>TV Camera Sync Generator</b>					μPC595		NEC
	TDA4280T/U	Siemens				CD22402		RCA		μPC596		NEC
	TDA4281T/U	Siemens			<b>TV Channel Selection—See also: Switches, Capacitance or Touch Sensing and TV Tuning, both this section</b>					AN5111		Panasonic
	TBA120U	Signetics		10						AN5120		Panasonic
	TBA750C	Signetics			<b>Universal Sync. Generator (NTSC)</b>				65	AN5710		Panasonic
	TDA2546A	Signetics				SCN2622A		Signetics		AN5720		Panasonic
	<b>TDA3810</b>	<b>Signetics (3613)</b>			<b>Universal Sync. Generator (PAL)</b>					SL1430		Plessey
	TDA1190Z	Sprague				SCN2621A		Signetics		SL1431		Plessey
	TDA3190	Sprague		15	<b>Universal Video Interface</b>					SL1432		Plessey
	ULN-2111A	Sprague			SCN2637A	Signetics		TDA440	Plessey			
	ULN-2290	Sprague		<b>Vertical Deflection</b>				CA1352	RCA			
	TBA120	Telefunken			MC1393	Motorola		CA270	RCA			
	SN76643	TI			MC6190	Motorola		CA3068	RCA			
	SN76651	TI	20		MC6191	Motorola		CA3136	RCA			
	SN76660	TI			MC6192	Motorola	70	CA3191	RCA			
	SN76665	TI			MC6193	Motorola		CA3192	RCA			
	SN76666	TI			MC6194	Motorola		CA3216	RCA			
	SN76688	TI			μPC1031	NEC		LA1352	Sanyo			
	TA7051	Toshiba	25		AN5510	Panasonic	75	LA1353	Sanyo			
	TA7072	Toshiba			AN5512	Panasonic		LA1354	Sanyo			
	TA7073A	Toshiba			AN5520	Panasonic		LA1357N	Sanyo			
	TA7130	Toshiba			LA1385	Sanyo		TDA4420	SGS			
	TA7146	Toshiba			LA1463	Sanyo		TDA4433	SGS			
	TA7176A	Toshiba	30		LA1464	Sanyo	80	SDA5500	Siemens			
	TA7243	Toshiba			LA7800	Sanyo		TBA1440G	Siemens			
	TA7314	Toshiba			TDA1170	SGS		TBA1441	Siemens			
	TA7632	Toshiba			TDA1270	SGS		TDA5500	Siemens			
	TA7633	Toshiba			TDA1470	SGS		TDA5600	Siemens			
					TDA1470A	SGS	85	TDA5610	Siemens			
<b>Surface Acoustic Wave Filter, IF</b>					TDA1670	SGS		TDA5611	Siemens			
	SW203	Plessey	35		TDA1770	SGS		TDA2544	Signetics			
<b>Synchron Generator</b>					TDA2653	Signetics		TDA3540	Signetics			
	2001Z	NCM			TDA2653A	Signetics		TDA3541	Signetics			
<b>Sync Generator</b>	SAA1043P	Signetics			TDA2655A	Signetics	90	TDA3564	Signetics			
<b>Synchronization and Deflection Circuit</b>					TDA2655B	Signetics		TDA440	Telefunken			
	LA7801	Sanyo			TDA3651	Signetics		TDA4400	Telefunken			
	LA7802	Sanyo			TDA3651A	Signetics		TDA4410	Telefunken			
	LA7806	Sanyo	40		TDA3652	Signetics		TDA4420	Telefunken			
	LA7810	Sanyo			TDA1170	Sprague	95	TDA4421	Telefunken			
<b>Synchronous Demodulator</b>					ULN-2270	Sprague		TDA4422	Telefunken			
	TDA0820T	Signetics			U2170B	Telefunken		TDA4430	Telefunken			
<b>Teletext, Channel and Page Selector</b>					TA7152	Toshiba		TDA4431	Telefunken			
	TEA9241	Thomson-CSF			TA7242	Toshiba		TDA4432	Telefunken			
<b>Teletext, DIDON Demodulator</b>							100	TDA4440	Telefunken			
	TEA2585	Thomson-CSF		<b>VHF Tuner</b>	TUA2000	Siemens		TDA4450	Telefunken			
	TEA2586	Thomson-CSF	45	<b>Video Circuits (IF amplifiers, detectors, AGC, AFC, and various combinations)</b>				SN76600	TI			
<b>Test Pattern Generator</b>					HA11215	Hitachi		SN76644	TI			
	ZNA234	Ferranti			HA11220	Hitachi		SN76650	TI			
<b>Timer, Programmable</b>					HA11221	Hitachi		TA7069	Toshiba			
	MM53100	National			HA11238	Hitachi	105	TA7074	Toshiba			
	MM53105	National			HA11405	Hitachi		TA7075	Toshiba			
<b>Tone Control</b>					MC13001	Motorola		TA7076	Toshiba			
	TDA1524	Signetics			MC13002	Motorola		TA7124	Toshiba			
	TA7125	Toshiba	50		MC13010	Motorola		TA7145	Toshiba			
<b>Touch Tuner: See Switches, Capacitance or touch sensing and TV</b>					MC1330A1	Motorola		TA7147	Toshiba			
					MC1330A2	Motorola	110	TA7162	Toshiba			
<b>Traffic Camera Controller</b>					MC1349	Motorola		TC5003	Toshiba			
	4004Z	NCM			MC1350	Motorola						
<b>Tuner Band Switching</b>					MC1352	Motorola						
	AN5700	Panasonic			MC1372	Motorola	115					
	LA7900	Sanyo			MC1373	Motorola		<b>Video Disc Player/Microprocessor Interface (digital auxiliary information buffer)</b>				
	TA7315	Toshiba	55		LM1822	National		CD3226	RCA			
<b>Tuner Controller</b>					LM1886	National		<b>Video Pulse Generator</b>				
	SL952	Plessey			LM1889	National		S178	Siemens			
								<b>Y Signal Recording Processor</b>				
								TA7637	Toshiba			

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Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Video Game Circuits</b>				<b>Motor Speed Regulator (Cont'd)</b>				<b>Switches (Cont'd)</b>			
Character Generator				MC2907	Cherry			SAS560	Siemens		
<b>MS3819</b>	<b>Harris</b>	<b>(4658)</b>		LM1014	Fairchild			SAS570	Siemens		105
Clock Generator (for MM57100)				LM1014A	Fairchild			SAS580	Siemens		
<b>MM53104</b>	<b>National</b>			μA7392	Fairchild		50	SAS590	Siemens		
Color Processor (generates single composite color signal from game chip input)				LS7263	LSIComp			SAS6810	Siemens		
<b>AY3-8915</b>	<b>GI</b>			LM1014	National			SAS660	Telefunken		
Dedicated (some may be used in programmable systems also)				<b>MSM5819</b>	<b>OKI</b>	<b>(4118)</b>		SAS6600	Telefunken		110
<b>AY3-8500</b>	<b>GI</b>		5	SL440	Plessey			SAS6610	Telefunken		
<b>AY3-8500-1</b>	<b>GI</b>			TDA1085A	Plessey		55	SAS670	Telefunken		
<b>AY3-8603</b>	<b>GI</b>			TDA2085A	Plessey			SAS6700	Telefunken		
<b>AY3-8605</b>	<b>GI</b>			TCA900	SGS			SAS6710	Telefunken		
<b>AY3-8606</b>	<b>GI</b>			TCA910	SGS						
<b>AY3-8607</b>	<b>GI</b>			TDA1151	SGS						
<b>AY3-8610</b>	<b>GI</b>		10	S572	Siemens						
Interface (processes graphics data and generates video signals)				TCA955	Siemens						
<b>AY3-8900</b>	<b>GI</b>			TDA1506	Signetics						
<b>AY3-8900-1</b>	<b>GI</b>			SG1731	SiliconG						
PAL Color Encoder and Video Summer				SG2731	SiliconG						
<b>TEA1002</b>	<b>Signetics</b>			SG3731	SiliconG						
Programmable Video Interface				Motor Speed Regulator, for Record Players, Tape Recorders	<b>TDA1533</b>	<b>Signetics</b>		Switches, Capacitive Touch Input (keyboard encoder and prioritizer)	<b>TMS1976</b>	<b>TI</b>	
<b>SCN2636A</b>	<b>Signetics</b>		15	Motor Speed Regulator, PLL	<b>TM4503</b>	<b>Toshiba</b>		Switches, DC-Controlled, Audio			115
<b>SCN2637A</b>	<b>Signetics</b>			Motor Speed Regulator, PLL, for Record Players, VTR	<b>MSM5818</b>	<b>OKI</b>	<b>(4118)</b>	LM1037	National		
Sound Generator				Motor, Stepping Motor Drive	<b>PBL3717</b>	<b>RIFA</b>		LM1038	National		
<b>AY3-8910</b>	<b>GI</b>			Motor, Stepping Motor Driver	<b>SAA1027</b>	<b>Signetics</b>		TDA1029	Signetics		
<b>AY3-8912</b>	<b>GI</b>			Music Chimes	<b>SAB0600</b>	<b>Siemens</b>					
<b>SN76477</b>	<b>TI</b>			Music, Voltage Controlled Oscillator	<b>SSM2030</b>	<b>SSM</b>		Switches, Proximity			120
TV/μP Interface				<b>SSM2033</b>	<b>SSM</b>			SAS221	Siemens		
<b>M106</b>	<b>SGS</b>			Music, Voltage Controlled Filter, Four Stage	<b>SSM2040</b>	<b>SSM</b>		TCA205AK	Siemens		
Video Display Generator				<b>SSM2044</b>	<b>SSM</b>			TDA0159	Thomson-CSF		
<b>MC6847</b>	<b>Motorola</b>	<b>(1505)</b>	20	Music, Voltage Controlled Transient Generator	<b>SSM2050</b>	<b>SSM</b>		TDA0161	Thomson-CSF		
Video Modulator (interfaces audio, color, difference, and luminance signals to TV antenna terminals)				<b>SSM2055</b>	<b>SSM</b>			TDA0162	Thomson-CSF		
<b>MC1372</b>	<b>Motorola</b>			<b>SSM2056</b>	<b>SSM</b>						
<b>MC1373</b>	<b>Motorola</b>			Optical Transceiver (intrusion alarm)	<b>CS258A</b>	<b>Cherry</b>		Switches, Touch-Sensitive Light Dimmer			125
<b>LM1889</b>	<b>National</b>			Peak Power Meter Driver, Dual (for hi-fi stereo)	<b>TA7318</b>	<b>Toshiba</b>		<b>LS7231</b>	<b>LSIComp</b>	<b>(803)</b>	
				<b>TA7332</b>	<b>Toshiba</b>			<b>LS7232</b>	<b>LSIComp</b>	<b>(803)</b>	
				Protector Circuit (overcurrent detection for OCL and speaker)	<b>HA12002</b>	<b>Hitachi</b>		<b>LS7233</b>	<b>LSIComp</b>	<b>(803)</b>	
				<b>TA7317</b>	<b>Toshiba</b>			<b>LS7234</b>	<b>LSIComp</b>	<b>(803)</b>	
				Revolution Counter Pulse Shaper	<b>SAK215</b>	<b>ITT</b>		<b>LS7235</b>	<b>LSIComp</b>	<b>(803)</b>	
				Servo Driver	<b>NE544</b>	<b>Signetics</b>	<b>(3632)</b>	S576A	Siemens		
				Smoke Detector				S576B	Siemens		
				<b>CS179</b>	<b>Cherry</b>			S576C	Siemens		130
				<b>MP7162</b>	<b>MicroPwr</b>			TEA1010	Signetics		
				<b>MC14466</b>	<b>Motorola</b>			TEA1058T	Signetics		
				<b>MC14467</b>	<b>Motorola</b>						
				<b>LM1801</b>	<b>National</b>			Tachometer Circuit			135
				<b>NF5301</b>	<b>National</b>			<b>CS189</b>	<b>Cherry</b>		
				<b>CA3164</b>	<b>RCA</b>	<b>(3604)</b>		<b>CS2917</b>	<b>Cherry</b>		
				<b>SCL5331-5</b>	<b>SSS</b>			<b>LM2907</b>	<b>National</b>		
				<b>SCL5331-6</b>	<b>SSS</b>			<b>LM2917</b>	<b>National</b>		
				<b>SCL5341</b>	<b>SSS</b>			<b>SN76810</b>	<b>TI</b>		
				<b>SD2</b>	<b>Supertex</b>						
				<b>SD3A</b>	<b>Supertex</b>			Thermo Controller (for air conditioners, refrigerators, etc.)	<b>TA7165</b>	<b>Toshiba</b>	
Switch, Video	<b>SAA1300</b>	<b>Signetics</b>						Timers. See Appliance Timers, Clock, Watch circuits in this section and Linear-Timers			
Switches	<b>SAS560</b>	<b>Hitachi</b>						Tone Synthesizer for Doorchimes, Toys	<b>AY3-1350</b>	<b>GI</b>	140
	<b>SAS570</b>	<b>Hitachi</b>						Traffic Light Control Circuit	<b>4004Z</b>	<b>NCM</b>	
	<b>MN1976</b>	<b>Panasonic</b>						Tuning Voltage Stabilizer, for use with Variable Capacitance Diodes	<b>TCA720</b>	<b>ITT</b>	
								<b>TAA550</b>	<b>SGS</b>		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## LINEAR-Followers

Bias Current nA 25°C	Offset Voltage mV 25°C	Voltage Drift μV/°C	Unity Gain Bandwidth MHz min	Slew Rate v/μs	Output V mA	Supply Range, V	Comments	Device	Source	Line
<b>Followers</b>										
See also Linear-Amplifiers, Special Purpose - Current Amplifiers										
0.1	10	25 *	100 *	1000	9 @ 100	± 5 to ± 20	Fast Follower	ADLH0033 LH0033 TP0033-80	† AD † National † TeledyneP	(3353)
	50	400	200	3000	10 @ 200	± 11 to ± 18	Very Fast, FET	9963	† OEI	5
0.15	20	25 *	100 *	1000	9 @ 100	± 5 to ± 20	Fast Follower	ADLH0033C LH0033C TP0033	AD National TeledyneP	(3353)
0.2	25	300 *	200 *	2000	10 @ 200	± 5 to ± 20	Very Fast Follower	LH0063	† National	10
	50	300 *	200 *	2000	10 @ 200	± 5 to ± 20	Very Fast Follower	LH0063C	National	
0.2 *	50	300 *	300 *	2000	10 @ 200	± 5 to ± 20	Gain 0.094 (50 Ohm)	3553	† Burr-Brown	
0.4	6.0	90	55 *	220	10 @ 22	± 6 to ± 18	Very Fast Follower	BUF-03E	PMI	
		100	55 *	220	10 @ 21	± 6 to ± 18	Very Fast Follower	BUF-03A	PMI	
0.7	15.0	150	50 *	180	10 @ 25	± 6 to ± 18	Very Fast Follower	BUF-03F	PMI	15
			170	50 *	180	± 6 to ± 18	Very Fast Follower	BUF-03B	† PMI	
3	4.0	6 *	12 *	30 *	10 @ 1	± 5 to ± 18	Follower, Replaces 102	SG210	SiliconG	
			20 *	30 *	10 @ 1	± 5 to ± 18	Dual 2 10 Follower	LH2210	National	
							Follower, Replaces 102	LM210	National	
								SFC2210	Thomson-CSF	
			10 *	20 *	30 *	10 @ 1	Follower, Replaces 102	LM110	† Intersil	20
			12 *	12 *	30 *	10 @ 1	Follower, Replaces 102	SG110	† SiliconG	
			20 *	30 *	10 @ 1	± 5 to ± 18	Dual 110 Follower	LH2110	Intersil	
								LH2110	† National	
							Follower, Replaces 102	LM110	† National	
5	20.0	100	200	1000	10 @ 500	± 11 to ± 18	Fast, High Current	9911	† OEI	25
7	7.5	10 *	12 *	30 *	10 @ 1	± 5 to ± 18	Follower, Replaces 102	SG310	SiliconG	
			20 *	30 *	10 @ 1	± 5 to ± 18	Dual 3 10 Follower	LH2310	Intersil	
								LH2310	† National	
							Follower, Replaces 102	LM310	Intersil	
								LM310	National	30
10	5.0	6 *	—	—	10 @ 1.25	± 15	Follower	LM102	† Intersil	
			8 *	10 *	10 @ 1	± 15	Follower	SG102	† SiliconG	
			10 *	10 *	10 @ 1.25	± 12 to ± 18	Follower	LM102	† National	
15	10.0	15 *	8 *	10 *	10 @ 1	± 15	Follower	SG202	SiliconG	
			10 *	10 *	10 @ 1.25	± 12 to ± 18	Follower	LM202	National	35
20	20.0	100	60	2000	10 @ 100	± 6 to ± 18	Very Fast Follower	9910	† OEI	
30	15.0	20 *	—	—	10 @ 1.25	± 15	Follower	LM302	Intersil	
			8 *	10 *	10 @ 1	± 15	Follower	SG302	SiliconG	
			10 *	10 *	10 @ 1.25	± 12 to ± 18	Follower	LM302	National	
	20	100	25	1000	10 @ ± 50	± 3 to ± 15	6 Identical Sections	MZ320	AnalogSys	40
35	15	33 *	250 *	1300	10 @ 100	± 5 to ± 16	Video Buffer	NA-5033-2	† Harris	(3462)
								NA-5033-5	Harris	(3462)
5000	5	25	125 *	1400	12 @ 100	—	Fast Follower	NOS-100A	AD	(3353)
								NOS-100S	† AD	(3353)
20000	20	100	24 *	1500 *	10 100	± 6 to ± 18	Follower, Current Booster	AH0010C	OEI	45
								AH0010F	OEI	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR  
Master Selection Guide

LINEAR-Operational Amplifiers

Function	Device	Source	Line	Function	Device	Source	Line	mA V	Device	Source	Line
<b>General Purpose</b>				FET Input. Bipolar/JFET available in standard (155), wideband (156) and uncompensated (157) versions, premium performance (A and B suffixes), and various temperature ranges (155, 255, 355, etc are available.) Some versions approach the low cost of the units below.				<b>High Output Current</b>			
Dual Amplifiers. The 1458 (0°C to 70°C) and the 1558 (-55°C to 125°C) offer approximately the same performance as the 741. Compensation is built in.								10000 @ 26 OPA501S † Burr-Brown			
XR1458 Exar								10000 @ 20 <b>OPA501</b> <b>Burr-Brown</b> (2850)			
μA1458 Fairchild				LF155 † AMD				OPA501A Burr-Brown			
MC1458 Motorola				LF155 † Intersil				5000 @ 45 PA07 Apex			
LM1458 National				LF155 † Motorola				5000 @ 42 PA10A † Apex			
<b>OP-14</b> <b>PMI</b> (3566) 5				PM-155 † PMI				5000 @ 37 PA10 Apex			
RC1458 Raytheon								PA10M † Apex			
CA1458 RCA								5000 @ 22 PA11 Apex			
<b>NE5512</b> <b>Signetics</b> (3638)								PA73 Apex			
<b>NE5514</b> <b>Signetics</b> (3640)								3500 @ 18 MCEL165 MCE			
<b>SE5512</b> † <b>Signetics</b> (526,3638) 10								MCEL2165 MCE			
<b>SE5514</b> † <b>Signetics</b> (3640)								3000 @ 18 MCETCA2365 MCE			
MC1458 SiliconG								MCETCA365 MCE			
SG1458 SiliconG								2700 @ 24 ICH8530I Intersil			
MC1458 TI								ICH8530M † Intersil			
								2000 @ 32 PA09 Apex			
								2000 @ 30 <b>3572A</b> <b>Burr-Brown</b> (2850)			
The 101A series consists of the 101A (-55°C to 125°C), the 201A (-25°C to 85°C) and the 301 (0°C to 70°C). These units require an external compensation capacitor which permits the bandwidth to be optimized for particular applications.								2000 @ 24 ICH8520I Intersil			
								ICH8520M † Intersil			
<b>AD101A</b> † <b>AD</b> (3351) 15								2000 @ 22 PA01 Apex			
μA101A Fairchild								LHO101 National			
LM101A † LinearTech								LHO101A † National			
LM101A † Motorola								LHO101AC National			
LM101A † National								LHO101C National			
LM101A † Raytheon								2000 @ 20 <b>3573</b> <b>Burr-Brown</b> (2850) 75			
<b>LM101</b> † <b>Signetics</b> (526) 20								2000 @ 18 LH0021 † National			
SG101A † SiliconG								2000 @ 10 CLC103A Comlinear			
LM101A † TI								CLC103AM † Comlinear			
TA7506 Toshiba								1500 @ 44 PA12A † Apex			
								1500 @ 34 PA12 Apex			
								PA12M † Apex			
								1000 @ 30 <b>3571A</b> <b>Burr-Brown</b> (2850) 80			
The performance of the 741 series is similar to the 101. These units include internal compensation to make the device stable and to eliminate the need for an external capacitor.								1000 @ 24 ICH8510I Intersil			
<b>AD741</b> † <b>AD</b> (3351) 25								ICH8510M † Intersil			
μA741 Fairchild								1000 @ 10 LH0021C National			
ICL741 † Intersil								750 @ 34 1461 TeledyneP			
<b>MP5502</b> † <b>MicroPwr</b> (3529)								500 @ 10 LH0061 † National			
MC1741 † Motorola								LH0061C National			
LM741 † National								200 @ 10 3553 † Burr-Brown			
μPC741 NEC								μA759 † Fairchild			
<b>OP-02</b> <b>PMI</b> (3566) 30								μA759C Fairchild			
PM-741 PMI								150 @ 140 PA08 Apex			
RC741 Raytheon								130 @ 13 LH0041 † National			
CA741 RCA								LH0041C National			
μA741 † Signetics								100 @ 10 AD3554A AD (3353)			
SG741 † SiliconG								AD3554S † AD (3353)			
SFC2741 Thomson-CSF								<b>H0S-060S</b> † <b>AD</b> (3353)			
TA7504 Toshiba								3329 † <b>Burr-Brown</b> (2850)			
								3554A <b>Burr-Brown</b> (2851)			
FET Input. The "741" of the FET input devices has yet to be established. Such a device requires very low cost, several sources, and many users. Commercial devices in the 155 series above meet some of these criteria, however, devices from the series below are generally lower in cost.								3554B <b>Burr-Brown</b> (2851)			
								3554S <b>Burr-Brown</b> (2851)			
μAF771 Fairchild								CLC200A Comlinear			
LF13741 National								CLC200AM † Comlinear			
LF351 National								LH0003 † National			
<b>CA3140</b> † <b>RCA</b> (3595) 40								TP3554 TeledyneP			
<b>CA3160</b> † <b>RCA</b> (3596) 45								100 @ - CA3094 † RCA			
<b>CA3260</b> <b>RCA</b> (3595)											
LF351 Motorola											

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers (Cont'd)

mA V	Device	Source	Line	Unity Gain Slew Rate (V/ $\mu$ s min)	Device	Source	Line	Maximum Supply Voltage	Device	Source	Line	
<b>High Output Current (Cont'd)</b>				<b>High Speed</b>				<b>High Voltage</b>				
75 @ 40	3583	Burr-Brown		8000	CLC220A	Comlinear		$\pm 150$	PA08	Apex		
75 @ 10	PA83	Apex			CLC220AM	† Comlinear			PA83	Apex		
	PA83M	† Apex		6000	CLC103A	Comlinear			PA83M	† Apex	105	
75 @ 145	1480	TeledyneP			CLC103AM	† Comlinear	50		PA84	Apex		
60 @ 30	<b>3580J</b>	<b>Burr-Brown</b>		4000	CLC200A	Comlinear			3582J	Burr-Brown		
		(2850)			CLC200AM	† Comlinear			3583	Burr-Brown		
50 @ 10	<b>AD380</b>	<b>AD</b>	5		CLC210A	Comlinear			3584	Burr-Brown		
	AD380S	† AD			CLC210AM	† Comlinear		$\pm 75$	3581J	Burr-Brown	110	
	<b>AD382</b>	<b>AD</b>		1500	ADLH0033	† AD			$\pm 50$	PA07	Apex	
	AD382S	† AD			HOS-100S	† AD			PA10A	† Apex		
	<b>AM-500</b>	<b>Datel (2861,2865)</b>	10	1400	ADLH0033C	AD			PA12A	† Apex		
	9914	OEI			HOS-100A	AD		$\pm 45$	PA10	Apex		
	9918	OEI		1000	<b>AD3554</b>	<b>AD (3353)</b>			PA12	Apex	115	
	AMP-01	† PMI			AD3554S	† AD		$\pm 40$	PA9	Apex		
	1430	TeledyneP			<b>3554</b>	<b>Burr-Brown</b>	60		3571A	Burr-Brown		
50 @ 5	LH0005	National	15			(2851)		3572A	Burr-Brown			
	LH0005A	† National			<b>AM-5006C</b>	<b>Datel (2861,2865)</b>			CLC210A	Comlinear		
47 @ 14	LH0020	† National			<b>AM-500MC</b>	<b>Datel (2861,2865)</b>			CLC210AM	† Comlinear	120	
	LH0020C	National			9914	OEI			<b>AM-464-2</b>	<b>Datel (2864)</b>		
40 @ 143	PA84	Apex		600	<b>HA-2539</b>	<b>Harris (3402)</b>	65		<b>AM-464-2M</b>	<b>† Datel (2864)</b>		
40 @ 4	LH0005C	National	20						<b>HA-2640</b>	<b>† Harris (3412)</b>		
30 @ 70	3581J	Burr-Brown		500	ADLH0032	† AD			HA-2645	Harris	125	
25 @ 10	<b>AM-430</b>	<b>Datel (2864)</b>			ADLH0032C	AD			LM143	† Harris		
	AM-430M	† Datel			1430	TeledyneP			MC1536	† Motorola		
20 @ 12	TAA2761	Siemens	25		1430-83	† TeledyneP			LH0004	† National		
	TAA2762	† Siemens		400	<b>HA-2540</b>	<b>Harris (3404)</b>	70		LH0004C	National		
	TAA2765	Siemens			LH0024	† National			LM143	† National		
	TAA4761	Siemens		350	LH0032	† National			LM144	† National	130	
	TAA4765	Siemens			LH0032C	National			LM1536	† National		
	TAA761	Siemens							SG1536	† SiliconG		
	TAA762	† Siemens							1332	TeledyneP		
	TCA311	Siemens	30					$\pm 35$	3580J	Burr-Brown		
	TCA312	† Siemens		300	HOS-050	† AD						
	TCA315	Siemens			HOS-050A	† AD		$\pm 34$	MC1436	Motorola	135	
	TCA325	Siemens			HOS-050C	AD			LM344	National		
	TCA331	Siemens	35		HOS-060S	† AD			SG1436	SiliconG		
	TCA332	† Siemens			9916	† OEI						
	TCA335	Siemens		280	HA2530	† Intersil						
20 @ 8	TAA861	Siemens	40					$\pm 30$	PA11	Apex		
	TAA862	† Siemens		250(g=5)	9906	† OEI	80		PA73	Apex		
	TAA865	Siemens							ICH8510I	Intersil	140	
15 @ 30	LH0004C	National		250	3551J	Burr-Brown			ICH8510M	† Intersil		
15 @ 10	<b>HA-2620</b>	<b>† Harris (3408)</b>			3551S	† Burr-Brown			ICH8520I	Intersil		
	HA2620	† Intersil			LH0024C	National			ICH8520M	† Intersil		
15 @ 145	3582J	Burr-Brown			1435	† TeledyneP			ICH8530I	Intersil		
	<b>3583</b>	<b>Burr-Brown</b>		200	AD380	AD			ICH8530M	† Intersil	145	
		(2850)			AD380S	† AD			MC1436C	Motorola		
	<b>3584</b>	<b>Burr-Brown</b>	45		<b>HA-5190</b>	<b>Harris (3455)</b>			LM1436	National		
		(2850)			HA-5195	Harris			LM343	National		
				100(g=3)	<b>HA-2520</b>	<b>† Harris (3400)</b>	90		SG1436C	SiliconG		
					HA2520	† Intersil						
				100	<b>AD509K</b>	<b>AD (3353)</b>						
					<b>AD509S</b>	<b>† AD (3353)</b>						
					<b>3550K</b>	<b>Burr-Brown</b>						
						(2851)						
					<b>OPA605</b>	<b>Burr-Brown</b>						
						(2851)						
					<b>HA-5160</b>	<b>Harris (3448)</b>	95					
					HA-5162	Harris						
				80	<b>AD509J</b>	<b>AD (3353)</b>						
					<b>3507J</b>	<b>Burr-Brown</b>						
						(2851)						
					1322	TeledyneP						
					1322-01	† TeledyneP	100					
				65	<b>3550J</b>	<b>Burr-Brown</b>						
						(2851)						
					<b>3550S</b>	<b>† Burr-Brown</b>						
						(2851)						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR

Master Selection Guide

LINEAR-Operational Amplifiers

Current pA 25°C	Device	Source	Line	Current pA 25°C	Device	Source	Line	Function	Device	Source	Line
<b>Low Bias Current</b>				(Cont'd)				<b>Low Drift</b>			
Rated values do not usually include thermal rise due to circuit operation.											
0.01	<b>ICH8500A</b>	<b>Intersil</b>	<b>(3477)</b>	5.0	1425-01	TeledyneP	60	0.5	AD510L	AD	
0.075	<b>3528C</b>	<b>Burr-Brown</b>			1425-02	TeledyneP			3510C	<b>Burr-Brown</b>	
	<b>AD515L</b>	<b>AD</b>	<b>(3351)</b>		CA3080	RCA			LT1002A	† <b>LinearTech</b>	
	<b>OPA104CM</b>	<b>Burr-Brown</b>			CA3080A	† <b>RCA</b>			LT1002AC	<b>LinearTech</b>	
			<b>(2850)</b>		<b>AD508L</b>	<b>AD</b>	<b>(3351)</b>		OP05A	† <b>LinearTech</b>	5
0.1	OPA106W	† <b>Burr-Brown</b>		10.0	OPA101BM	<b>Burr-Brown</b>			LH0044B	<b>National</b>	
	AM-490-2C	<b>Datel</b>			OPA102BM	<b>Burr-Brown</b>			OP-05A	† <b>PMI</b>	
	3523L	<b>Burr-Brown</b>			<b>OPA106/MIL</b>	† <b>Burr-Brown</b>					
	<b>ICH8500</b>	<b>Intersil</b>	<b>(3477)</b>		<b>MP5507A</b>	† <b>MicroPwr</b>	<b>(3529)</b>	0.6	OPA27A	† <b>Burr-Brown</b>	
0.15	3528B	<b>Burr-Brown</b>	10						OPA27E	<b>Burr-Brown</b>	
	OPA106V	† <b>Burr-Brown</b>							OPA37A	† <b>Burr-Brown</b>	10
	<b>AD515K</b>	<b>AD</b>	<b>(3351)</b>						OPA37E	<b>Burr-Brown</b>	
	<b>OPA104BM</b>	<b>Burr-Brown</b>	<b>(2850)</b>						AM-430B	<b>Datel</b>	
0.2	AD0P-07	† <b>AD</b>							AM-490-2M	† <b>Datel</b>	
	<b>HA-5180</b>	<b>Harris</b>	<b>(3452)</b>						μA714	† <b>Fairchild</b>	
	<b>HA-5180A</b>	<b>Harris</b>	<b>(3452)</b>						HA-5130-5	<b>Harris</b>	15
0.25	AD523L	<b>AD</b>							HA-5135-5	<b>Harris</b>	
	3523K	<b>Burr-Brown</b>							HA-OP07	<b>Harris</b>	
0.3	3528A	<b>Burr-Brown</b>							LT1001AC	<b>LinearTech</b>	
	<b>OPA103</b>	<b>Burr-Brown</b>	<b>(2850)</b>						LT1001AM	† <b>LinearTech</b>	20
	OPA106U	† <b>Burr-Brown</b>							LT1007AC	<b>LinearTech</b>	
	AM-490-2B	<b>Datel</b>							LT1007AM	† <b>LinearTech</b>	
	<b>AD515J</b>	<b>AD</b>	<b>(3351)</b>						LT1037AC	<b>LinearTech</b>	
	<b>OPA104AM</b>	<b>Burr-Brown</b>	<b>(2850)</b>						LT1037AM	† <b>LinearTech</b>	
0.5	AD504M	<b>AD</b>	25						OP05E	<b>LinearTech</b>	
	AD510L	<b>AD</b>							OP07A	† <b>LinearTech</b>	25
	AD517L	<b>AD</b>							OP27A	† <b>LinearTech</b>	
	3510C	<b>Burr-Brown</b>							OP27E	<b>LinearTech</b>	
	MP517L	<b>MicroPwr</b>							OP37A	† <b>LinearTech</b>	
	MP5505A	† <b>MicroPwr</b>	30						OP37E	<b>LinearTech</b>	
	LH0044A	† <b>National</b>							OP27A	† <b>Motorola</b>	30
	LH0044AC	<b>National</b>							OP27E	<b>Motorola</b>	
	AD523K	<b>AD</b>							OP37A	† <b>Motorola</b>	
	3523J	<b>Burr-Brown</b>							OP37E	<b>Motorola</b>	
1.0	AD545K	<b>AD</b>	35						OP-05E	<b>PMI</b>	
	AD545L	<b>AD</b>							OP-07A	† <b>PMI</b>	35
	AD545M	<b>AD</b>							MP5507A	† <b>MicroPwr</b>	
	<b>OPA100</b>	<b>Burr-Brown</b>	<b>(2850)</b>						<b>MP5507E</b>	<b>MicroPwr</b>	<b>(3529)</b>
	OPA100C	<b>Burr-Brown</b>							1.0	AD504L	<b>AD</b>
	OPA103	<b>Burr-Brown</b>	40							AD504S	† <b>AD</b>
	<b>OPA105</b>	† <b>Burr-Brown</b>	<b>(2853)</b>							AD517K	<b>AD</b>
	AD523J	<b>AD</b>								AD517S	† <b>AD</b>
	<b>3522L</b>	<b>Burr-Brown</b>	<b>(2850)</b>							AD547L	<b>AD</b>
	ICL8007AC	<b>Intersil</b>								<b>3291</b>	<b>Burr-Brown</b>
	ICL8007AM	† <b>Intersil</b>	45								<b>(2850)</b>
	LH0052	† <b>National</b>								3500E	<b>Burr-Brown</b>
2.0	OPA100	<b>Burr-Brown</b>								3510B	<b>Burr-Brown</b>
	OPA100B	<b>Burr-Brown</b>								3510S	† <b>Burr-Brown</b>
	<b>3527B</b>	<b>Burr-Brown</b>	<b>(2850)</b>							3521L	<b>Burr-Brown</b>
	<b>OPA103BM</b>	<b>Burr-Brown</b>	<b>(2850)</b>							AM-490-2A	<b>Datel</b>
										μA725A	† <b>Fairchild</b>
										μA725E	<b>Fairchild</b>
										OP05	† <b>LinearTech</b>
										MP517K	<b>MicroPwr</b>
										MP517S	† <b>MicroPwr</b>
										MP5505A	† <b>MicroPwr</b>
										LH0044	† <b>National</b>
										LH0044C	<b>National</b>
										LM11	† <b>National</b>
										LM11C	<b>National</b>
										OP-05	† <b>PMI</b>
										<b>AD510K</b>	<b>AD</b>
										<b>AD510S</b>	† <b>AD</b>
										LM725A	<b>National</b>
										1.0 (dual unit)	<b>MP5510A</b>
											† <b>MicroPwr</b>
											<b>MP5510B</b>
											<b>MicroPwr</b>
											<b>MP5510E</b>
											<b>MicroPwr</b>
											OP-10
											† <b>PMI</b>
											OP-10A
											† <b>PMI</b>
											PMI
											1.3
											AM-430M
											† <b>Datel</b>

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## LINEAR-Operational Amplifiers (Cont'd)

Supply Current $\mu\text{A} \pm 15\text{V}$	Device	Source	Line	Supply Current $\mu\text{A} \pm 15\text{V}$	Device	Source	Line	Supply Current $\mu\text{A} \pm 15\text{V}$	Device	Source	Line
<b>Low Power Non-Adjustable</b>				<b>Low Power Programmable</b>				<b>Low Power <math>\mu\text{A} @ \pm 8\text{V}</math></b>			
9	OP-215B	† PMI	(3566)	70	The current drain limits of these units are approximately 10-300 $\mu\text{A}$ .			20	ICL7611M	† Intersil	(3481)
	OP-215C	† PMI	(3566)		$\mu\text{A}776$	† Fairchild			ICL7631M	† Intersil	(3481)
	OP-215F	PMI	(3566)		LM4250	† Harris			ICL7632M	† Intersil	(3481)
	OP-215G	PMI	(3566)		ICL8021C	Intersil	(3516)		ICL7642M	† Intersil	(3481)
45	OP-20B	† PMI		75	ICL8021M	† Intersil	(3516)	200	ICL7612M	† Intersil	(3481)
	OP-20C	† PMI			MC1776C	Motorola			ICL7613C	Intersil	(3481)
	OP-20F	PMI			MC3476	Motorola			ICL7613M	† Intersil	(3481)
	OP-20G	PMI			CA3078A	† RCA	(3597)		ICL7614C	Intersil	(3481)
	OP-20H	PMI			CA6078A	† RCA			ICL7615C	Intersil	(3481)
65	HA-5141A	Harris		80	SG1250	† SiliconG		125	ICL7621C	Intersil	(3481)
	HA-5142A	Harris			SG2250	SiliconG			ICL7622C	Intersil	(3481)
	HA-5144A	Harris			CA3094	† RCA			ICL7631C	Intersil	(3481)
80	HA-5141	Harris			CA3094A	† RCA	(3598)		ICL7642C	Intersil	(3481)
	HA-5142	Harris			CA3094B	† RCA	(3598)		ICL7612C	Intersil	(3481)
	HA-5144	Harris			The current drain limits of these units are approximately 10-300 $\mu\text{A}$ .				ICL7611C	Intersil	(3481)
200	TL061 Series	TI		85	$\mu\text{A}776\text{C}$	Fairchild		130			
220	OP-420B	† PMI	(3566)		HA-2720	† Harris	(3416)				
	OP-420C	† PMI	(3566)		HA-2725	Harris					
	OP-420F	PMI	(3566)		LM4250C	Harris					
	OP-420G	PMI	(3566)		ICL4250	† Intersil					
230	OP21A	Burr-Brown		90	ICL4250C	† Intersil		135			
	OP21E	Burr-Brown			MC1776	† Motorola					
250	OP21B	Burr-Brown			LM4250	† National					
	OP21F	Burr-Brown			LM4250C	† National					
275	OP21G	Burr-Brown			SG3250	SiliconG		140			
350	OP-21B	† PMI		95	SG4250	† SiliconG					
	OP-21F	PMI			SG4250C	† SiliconG					
400	OP-21G	PMI			CA3078	† RCA	(3597)				
	LF442A	National			UC4250	† Solitron					
500	LF442	National			UC4250C	Solitron					
600	$\mu\text{A}108$	† Fairchild		100	The current drain limits of these units are approximately 1-2000 $\mu\text{A}$ .						
	MP5512	† MicroPwr			OP-32	PMI	(3566)				
	OP-08	† PMI	(3566)								
	OP-12	† PMI	(3566)								
	LM112	† National		105							
	LM212	National									
	LM316A	National									
	LM108	† AMD									
	AD108	† AD	(3351)								
	LM108	† Harris		110							
	LM108	† Intersil									
	LM108	† Motorola									
	LM108	† National									
	PM-108	† PMI	(3566)								
	SG108	† SiliconG		115							
	SFC2108	† Thomson-CSF									

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR Master Selection Guide

LINEAR-Operational Amplifiers (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Supply Voltage, V	Device	Source	Line	
<b>Programmable</b>				Triple (3 units per package) (Cont'd)				<b>Single Supply</b>				
Adjustable Current/Performance				L144A † Siliconix (3083)				<b>Single Units</b>				
Single (1 unit per package)				L144B Siliconix (3083)				to 24 CA3094 † RCA				
LM4250 † Harris				L144C Siliconix (3083)				to 30 OP-32 PMI				
ICL7611M † Intersil				CA3080 † RCA (3597)				to 36 μA759 † Fairchild				
ICL7612M † Intersil				Quad (4 units per package)				μA759C † Fairchild				
ICL7613C Intersil				XR246 Exar (3377)				CA3140 † RCA				
ICL7613M † Intersil				XR346 Exar (3377)				CA3094A † RCA				
ICL8021C Intersil				XR4202 Exar (3378)				to 44 CA3094B † RCA				
ICL8021M † Intersil				XR4202M † Exar (3378)				1 to 16 ICL7612M † Intersil				
MC1776C Motorola				HA-2740 Harris (3422)				ICL7612C Intersil				
MC3476 Motorola				MC3575 † Motorola				2 to 30 HA-5141 Harris				
CA3078A † RCA				LM146 † National				5 to 16 OP-20B † PMI				
SG1250 † SiliconG				LM246 National				OP-20C † PMI				
SG2250 SiliconG				LM346 National				OP-20F PMI				
ICL7612C Intersil				LM146 Raytheon				OP-20G PMI				
CA3094 † RCA				LM246 Raytheon				OP-20H PMI				
CA3094A † RCA				LM346 Raytheon				CA3130 † RCA (3596)				
CA3094B † RCA				XR146 † Exar (3377)				CA3130A † RCA (3596)				
μA776C Fairchild								Dual Units				
μA776M † Fairchild								(Continued on next page)				
HA-2720 † Harris												
HA-2725 Harris												
LM4250C Harris												
ICL4250 † Intersil												
ICL4250C † Intersil												
ICL7611C Intersil												
MC1776 † Motorola												
LM4250 † National												
LM4250C † National												
CA3080 RCA												
CA3080A † RCA												
SG3250 SiliconG												
SG4250 † SiliconG												
SG4250C † SiliconG												
CA3078 † RCA												
Dual (2 units per package)												
HA-2730 † Harris (3419)												
HA-2735 Harris												
ICL8022C Intersil (3516)												
ICL8022M † Intersil (3516)												
LH24250 † National												
LH24250C National												
Triple (3 units per package)												
ICL7632C Intersil (3481)												
ICL7632M † Intersil												
ICL8023C Intersil (3516)												
ICL8023M † Intersil (3516)												
CA3060A RCA (3597)												
CA3060B RCA (3597)												
SG1253 † SiliconG												
SG2253 SiliconG												
SG3253 SiliconG												
				(Continued)								

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR—Operational Amplifiers (Cont'd)

Supply Voltage, V	Device	Source	Line	Supply Voltage, V	Device	Source	Line	Unity Gain Bandwidth MHz	Device	Source	Line
<b>Single Supply</b>				<b>Single Supply</b>				<b>Wide Band</b>			
<b>Dual Units</b>				<b>Quad Units</b>							
to 20	OP-215B	† PMI		2 to 30	HA-5144	Harris		40 *	AD380J	AD (3353)	
	OP-215C	† PMI		3 to 26	TDF2902	Thomson-CSF			AD380K	AD	105
	OP-215F	PMI		3 to 30	μA124	† Fairchild	45		AD380L	AD	
	OP-215G	PMI			μA224	Fairchild			AD380S	† AD	
2 to 30	HA-5142	Harris	5		μA324	Fairchild			HA-2400	† Harris	
3 to 26	HA17902	Hitachi			LM124	† Intersil			HA-2404	Harris	
	LM2902	Motorola			LM324	Intersil			HA-2405	Harris	
	LM2904	Motorola			LM124	† Motorola		50 *	3551J	Burr-Brown	110
	LM2902	National	10		LM224	Motorola			3551S	† Burr-Brown	
	LM2904	National			LM324	Motorola			CA3015	† RCA	
	LM2902	Raytheon			LM124	† National			CA3015A	† RCA	
	CA2904	RCA			LM224	National			CA3030	RCA	115
	LM2902	TI			LM324	National			CA3030A	RCA	
	LM2904	TI			μPC324	NEC			CA3038	† RCA	
	TDF2904	Thomson-CSF	15		LM124	† Raytheon		60 *	1430	TeledyneP	
3 to 30	LM158A	† National			LM224	Raytheon			1430-83	† TeledyneP	120
	LM358	National			LM324	Raytheon					
	CA358	RCA			CA124	† RCA		65 *	μA715C	Fairchild	
	LM358	Signetics			CA224	RCA			μA715M	† Fairchild	
	LM358	TI	20		CA324	RCA		70	HA-5112	Harris (3459)	
	LM158	† Motorola			LM124	† Signetics			HA-5114	Harris (3459)	
	LM258	Motorola			LM224	Signetics		70 *	OPA605	Burr-Brown	125
	LM358	Motorola			LM324	Signetics			3554A	Burr-Brown	
	LM158	† National	25		SG124	† SiliconG			3554B	Burr-Brown	
	LM258	National			SG224	SiliconG			3554S	Burr-Brown	
	LM258A	National			SG324	SiliconG			LH0032	† National	
	LM358A	National			LM124	† TI			LH0032C	National	
	CA158	† RCA			LM224	TI			ADLH0032	† AD (3353)	130
	CA258	RCA			LM258	TI			ADLH0032C	AD (3353)	
	LM158	† Signetics	30		LM324	TI			LH0024	† National	
	LM258	Signetics							LH0024C	National	
	NE532	Signetics		3 to 36	XR3403	Exar (3378)		70 (-3db)	CLC210A	Comlinear	
	SE532	† Signetics (526)			XR3503	† Exar (3378)			CLC210AM	† Comlinear	135
	LM158	† TI			μA3303C	Fairchild		90 *	AD3554	AD	
	TL321C	TI	35		μA3403C	Fairchild			AD3554S	† AD	
	TL321M	† TI			MC3303	Motorola		100	AM-5006C	Datel	
3 to 36	μA798	Fairchild			MC3403	Motorola			AM-500MC	Datel	
	MC3458	Motorola			MC3503	† Motorola			HA-5160	Harris	140
4 to 36	CA3240	RCA (3595)			RC3403	Raytheon			HA-5162	Harris	
5 to 30	OP-220	PMI (3566)			RM3503	† Raytheon			9906	† OEI	
	OP-221	PMI	40	3 to 44	MC33074	Motorola			NE5539	Signetics (3678)	
8 to 36	TBA231	SGS			MC34074	Motorola			SE5539	† Signetics (526,3678)	
					MC35074	† Motorola		85			
				4 to 18	MC3401	Motorola		100 *	HOS-050	† AD	145
					LM3401	National			HOS-050C	AD	
					RC3401	Raytheon			HOS-060S	† AD	
					CA3401	† RCA			HA-2620	† Harris	
				4 to 28	MC3301	Motorola			HA-2622	† Harris	
					LM3301	National		90	HA-2625	Harris	150
					RC3301	Raytheon					
				4 to 32	LM2900	National		100 (-3db)	CLC200A	Comlinear	
					LM3900	National			CLC200AM	† Comlinear	
					LM2900	Raytheon		150 (-3db)	CLC103A	Comlinear	
					LM3900	Raytheon			CLC103AM	† Comlinear	
					LM2900	TI		200	9918	OEI	155
					LM3900	TI			9916	† OEI	
				4 to 36	LM1900	TI		200 (-3db)	CLC220A	Comlinear	
5 to 30	OP-420B	† PMI							CLC220AM	† Comlinear	
	OP-420C	† PMI						300	9914	OEI	
	OP-420F	PMI						400 *	HA-2540	Harris	160
	OP-420G	PMI						500 *	1435	† TeledyneP	
								600 *	HA-2539	Harris	
								1000	OPA600/MIL	Burr-Brown (2853)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR—Operational Amplifiers—Characteristics

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>												
In this list the "Comp." column indicates the number of compensation components required at unity gain. Thus 0 indicates a fully compensated amplifier. The bandwidth and slew rate are listed with the amplifier compensated for unity gain.												
0.005	0.01	—	—	0.3 *	1 *	10M	110	0	Chopper Stabilized	TML7650-1	Telmos	5
				2.0 *	2.5 *	10M	120	0	Chopper Stabilized	TML7650	Telmos	
		0.0005	0.05	2	2.5	10M	120	0	Chopper Stabilized	<b>TSC7650</b>	<b>TeledyneS (3096)</b>	
		0.0005 *	0.05	2 *	2.5 *	5M	130 *	1	Chopper Stabilized	<b>AM-7650</b>	<b>Datel (2865)</b>	
		0.005	0.01	2.0 *	2.5 *	10M	120	0	Chopper Stabilized	<b>ICL7650C</b>	<b>Intersil (3488)</b>	
	3	—	0.1	0.3	—	30K	88 *	0	Commutating Auto Zero, Compensated	ICL7600C	Intersil	10
									Commutating Auto Zero, Uncompensated	ICL7600M	† Intersil	
										ICL7601C	Intersil	
										ICL7601M	† Intersil	
0.01 *	10	—	0.2 *	8 *	2.8 *	1.5M *	120	0	Ultra-Low Noise	<b>MP5527</b>	<b>MicroPwr (3529)</b>	10
				63 *	17 *	1.5M *	120	0	Ultra-Low Noise	<b>MP5537</b>	<b>MicroPwr (3529)</b>	
0.015	1	—	0.1	—	—	—	140	0	Instrumentation, 50 mA output	<b>AMP-01</b>	† PMI (3568)	15
	2	2	0.6	0.4	0.1	450K	114	0	Ultra Low Offset, Low Drift	LT1001AM	† LinearTech	
0.020 *	0.15 *	0.05 *	0.1	3 *	2.5 *	1M	120	3	Chopper Stabilized	<b>AM-490-2C</b>	<b>Datel (2865)</b>	15
			0.3	3 *	2.5 *	1M	120	3	Chopper Stabilized	<b>AM-490-2B</b>	<b>Datel (2865)</b>	
			0.6	3 *	2.5 *	1M	120	3	Chopper Stabilized	<b>AM-490-2M</b>	† <b>Datel (2865)</b>	
			1.0	3 *	2.5 *	1M	120	3	Chopper Stabilized	<b>AM-490-2A</b>	<b>Datel (2865)</b>	
0.025	1	0.25	0.5	0.25 *	0.1 *	1M	110	0	Ultra Low Offset Voltage, Low Drift	<b>AD517L</b>	<b>AD (3351)</b>	20
	2	2	0.6	0.4	0.1	450K	114	0	Ultra Low Offset, Low Drift	LT1001AC	LinearTech	
				0.6 *	0.17 *	300K	110	0	Ultra Low Offset Voltage, Low Drift	<b>ADOP-07A</b>	† <b>AD (3351)</b>	
										OP-07A	† Intersil	
				0.5	—	1M	110	0	Ultra Low Offset Voltage, Low Drift	HA-OP07A	Harris	25
										HA-OP07E	Harris	
				1.2 *	0.25 *	300K	110	0	Ultra Low Offset Voltage, Low Drift	<b>MP5507A</b>	† MicroPwr	
										<b>OP-07A</b>	† <b>MicroPwr (3529)</b>	
										<b>OP-07A</b>	† <b>PMI (3566)</b>	
				2.5 *	0.8 *	1M	100	0	Ultra Low Offset Voltage, Low Drift	<b>HA-5130-2</b>	<b>Harris (3444)</b>	30
										HA-5135-2	Harris	
										<b>HA-5130-5</b>	<b>Harris (3444)</b>	
										HA-5135-5	Harris	
	10	2.5	2	0.3 *	0.1 *	1M	110	0	Trimmed Offset	<b>AD510L</b>	<b>AD (3351)</b>	35
	15	2.5	0.5	0.4 *	0.06 *	1M	120	1	Precision, Low Noise	LH0044A	† National	
										LH0044AC	National	
	20	20	0.75	—	—	5-1000	100	0	Instrumentation (2 amps available as pair)	<b>3629B</b>	† <b>Burr-Brown (2851)</b>	40
	25	30	0.5	—	—	5-1000	100	0	Instrumentation (2 amps available as pair)	<b>3629C</b>	<b>Burr-Brown (2851)</b>	
										<b>3629S</b>	† <b>Burr-Brown (2851)</b>	
	35	50	3	—	—	5-1000	100	0	Instrumentation (2 amps available as pair)	<b>3629A</b>	<b>Burr-Brown (2851)</b>	45
40	35	—	—	63 *	11 *	—	—	0	Low Noise, High Slew Rate	<b>OP-37A</b>	† <b>Raytheon (3592)</b>	
				25 *	18 *	1M	—	0	Ultra Low Noise, Low Offset	<b>OP-37A</b>	† <b>PMI (3568)</b>	
										<b>OP-37E</b>	<b>PMI (3566)</b>	
			0.6	5	1.7	1M	114	0	Ultra Low Noise	<b>AM-427-B</b>	<b>Datel (2864)</b>	50
										OP-27A	LinearTech	
										OP-27E	LinearTech	
										OP-27A	† Motorola	
										OP-27E	Motorola	
										<b>OP-27A</b>	† <b>PMI (3566)</b>	
										<b>OP-27E</b>	<b>PMI (3566)</b>	
						600K	114	0	Ultra Low Noise	<b>OP-27A</b>	† <b>Raytheon (3592)</b>	
										<b>OP-27E</b>	<b>Raytheon (3592)</b>	
				8 *	1.7	1000K	114	0	Ultra Low Noise	<b>OPA27A</b>	† <b>Burr-Brown (2850)</b>	55
										<b>OPA27E</b>	† <b>Burr-Brown (2850)</b>	
				40 *	11	1000K	114	0	Ultra Low Noise	<b>OPA37A</b>	† <b>Burr-Brown (2850)</b>	
										<b>OPA37E</b>	<b>Burr-Brown (2850)</b>	
				45	11	1M	114	0	Ultra Low Noise, Gain >5	OP-37A	† LinearTech	
										OP-37E	LinearTech	
										OP-37A	† Motorola	
										OP-37E	Motorola	

† Military Temperature Range (−55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR—Operational Amplifiers—Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
0.05	0.014	0.014	45	7	13 *	128	86	0	High Speed Instrumentation	AM-543MM AM-543MR	† Datel Datel	
0.05	0.05	0.05	45	7	13 *	128	86	0	High Speed Instrumentation	AM-543MC	Datel	
2	0.75	1	0.25 *	0.1 *	1M	100	0	0	Ultra Low Offset Voltage, Low Drift	AD517K AD517S MP517K MP517S	AD † AD MicroPwr † MicroPwr	(3351) (3351)
5	0.5	0.5	0.2	—	2M	126	0	0	Precision Instrumentation	LM163A LM363A	† National † National	
13	4.0	1	0.3 *	0.1 *	1M	110	0	0	Trimmed Offset	AD510K AD510S	AD † AD	10
14	14	1	0.5 *	—	1024	86	0	0	Low Drift Instrumentation	AM-542MM AM-542MR	† Datel Datel	
15	10	0.5	—	5	1000	90	0	0	Precision Instrumentation	AD524C	AD	(3354)
30	—	0.2	0.25	5	1K	115	0	0	Precision Instrumentation	AD624C	AD	(3355)
5	0.5	0.4 *	0.06 *	500K	114	1	1	1	Precision, Low Noise	LH0044B	National	
1	0.4 *	0.06 *	500K	114	1	1	1	1	Precision, Low Noise	LH0044	† National	
50	50	1	0.5 *	—	1024	86	0	0	Low Drift Instrumentation	AM-542MC	Datel	
0.06	4	3.8	1	0.4	0.1	400K	110	0	Low Offset Voltage	LT1001C LT1001M	LinearTech † LinearTech	20
15	10	0.5	0.4 *	0.5 *	1M	110	1	1	Low Drift	3510C	Burr-Brown (2850,2853)	
55	50	—	63 *	11 *	—	—	0	0	Low Noise, High Slew Rate	OP-37B OP-37F	† Raytheon Raytheon	(3592) (3592)
1.3	5	1.7	1M	106	0	0	0	0	Ultra Low Noise	OPA27B OPA27F OP-27B OP-27F OP-27B OP-27F OP-27B OP-27F	† Burr-Brown Burr-Brown † Motorola Motorola PMI PMI † Raytheon Raytheon	(2850) (2850) (3566) (3566) (3592) (3592)
40 *	11	1000K	106	0	0	0	0	0	Ultra Low Noise	OPA37B OPA37F	† Burr-Brown Burr-Brown	(2850) (2850)
45	11	1M	106	0	0	0	0	0	Ultra Low Noise, Gain >5	OP-37B OP-37F OP-37B OP-37F	† Motorola Motorola † PMI PMI	(3566) (3566)
0.075	3	2.8	1.3	0.4	0.1	200K	110	0	Low Offset Voltage	OP-07	† LinearTech	
0.6 *	0.17 *	200K	110	0	0	0	0	0	Ultra Low Offset Voltage, Low Drift	OP-07 ADOP-07	† Intersil † AD	(3351)
0.5	1M	110	0	0	0	0	0	0	Ultra Low Offset Voltage, Low Drift	HA-0P07	Harris	
1.2 *	0.25 *	200K	110	0	0	0	0	0	Ultra Low Offset Voltage, Low Drift	MP5507B OP-07 OP-07 RC714 RC714C RC714E μA714	† MicroPwr † MicroPwr † PMI † Raytheon Raytheon Raytheon † Fairchild	(3529) (3529) (3566) (3592) (3592) (3592)
4	3.8	0.2	0.4	0.2	150K	106	0	0	Ultra-Low Offset Voltage	OP-07E	TI	
1.3	0.4	0.1	200K	110	0	0	0	0	Low Offset Voltage	OP-07E	LinearTech	
0.6 *	0.17 *	200K	123 *	0	0	0	0	0	Ultra Low Offset Voltage, Low Drift	ADOP-07E	AD	(3351)
1.2 *	0.25 *	200K	106	0	0	0	0	0	Ultra Low Offset Voltage, Low Drift	OP-07E μA714E MP5507E OP-07E OP-07E	Intersil Fairchild MicroPwr MicroPwr PMI	(3529) (3566)
6	1.3	0.6 *	0.5	1M	106	0	0	0	Ultra Low Offset Voltage, Low Drift	HA-0P07C	Harris	
15	3	2	1.2 *	0.25 *	1M	120	0	0	Precision Instrumentation	CA3193B CA3493B	† RCA † RCA	(3594) (3594)
0.08	0.050	0.01	1.5	1 *	0.3 *	500K	114	0	Improved LM11	LM11A	† National	
1	0.5	0.2 *	3 *	2.5 *	1M	120	3	3	Chopper Stabilized	AM-490 1340	Datel TeledyneP	(2865)
15	0.5	1.5	0.1 *	—	1M	105	0	0	Precision, Low Power	OP-20B OP-20F	† PMI PMI	(3566) (3566)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift µV/°C	Bandwidth MHz	Slew Rate V/µs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Single Units</b>										<b>(Cont'd)</b>			
0.10	3	2.8	0.4 *	1.2 *	0.25 *	500K	—	0	Ultra Low Offset Voltage	OP-207A OP-207B	† PMI PMI	(3566) (3566)	
	5	1	3	0.2 *	—	2M	120	0	Precision Instrumentation	LM163	† National		
	20	15	0.5	—	5	1000	90	0	Precision Instrumentation	AD524B	AD	(3354)	
	25	0.7	1	0.5 *	49	1000K	95	1	Selectable Input	OPA201C	Burr-Brown	(2850)	
		1	1	0.3 *	0.2	1M	100	0	Low Power	OPA21A OPA21E	Burr-Brown Burr-Brown	(2850) (2850)	
		5	3	3 *	0.1 *	250K	94	0	Trimmed Offset	AD510J	AD	(3351)	
	35	5	1	0.4 *	0.06 *	500K	114	1	Precision, Low Noise	LH0044C	National		
	40	1	0.8	—	1.7	1800	114	0	Ultra Low Noise	OP-27	MicroPwr	(3529)	
	50	—	0.5	0.25	5	1K	110	0	Precision Instrumentation	AD624B	AD	(3355)	
		35	0.5	—	5	1000	90	0	Precision Instrumentation	AD524S	† AD	(3354)	
	70	1	0.8	—	0.01	100K	120	4	High Gain Instrumentation	OP-06A	† PMI	(3566)	
	80	40	0.4 *	8 *	17	1800	114	0	Low Noise, High Slew Rate	OP-37	MicroPwr	(3529)	
		75	—	63 *	11 *	—	—	0		OP-37C	† Raytheon	(3592)	
								0		OP-37G	Raytheon	(3592)	
			0.4 *	8 *	3.2 *	1200K	120	0	Ultra Low Noise	OP-27G	Motorola		
						1200K *	120	0	Ultra Low Noise	OP-27C	Motorola		
				25 *	18 *	700K	120	0	Ultra Low Noise	OP-37C	† PMI	(3566)	
										OP-37G	PMI	(3566)	
		1.8	5	1.7	200K	100	0	0	Ultra Low Noise	AM-427-A	Datel	(2864)	
										AM-427-M	† Datel	(2864)	
					300K	100	0	0	Ultra Low Noise	OP-27C	† Raytheon	(3592)	
										OP-27G	Raytheon	(3592)	
					700K	100	0	0	Ultra Low Noise	OP-27C	† LinearTech		
										OP-27G	LinearTech		
				8 *	1.7	700K	100	0	Ultra Low Noise	OPA27C	† Burr-Brown	(2850)	
										OPA27G	Burr-Brown	(2850)	
										OPA37C	† Burr-Brown	(2850)	
										OPA37G	Burr-Brown	(2850)	
				45	11	700K	100	0	Ultra Low Noise, Gain > 5	OP-37C	† LinearTech		
										OP-37G	LinearTech		
									Ultra Low Noise, Gain > 5	OP-37C	† Motorola		
										OP-37G	Motorola		
	100	—	1	0.25	5	1K	100	0	Precision Instrumentation	AD624S	† AD	(3355)	
	150	15	0.5	0.6 *	0.25	2M	110	0	Precision, Low Power	OP-21B	† PMI	(3566)	
										OP-21F	PMI	(3566)	
0.12	25	15	1	0.4 *	0.5	1M	110	1	Low Drift	3510B	Burr-Brown	(2850,2853)	
										3510S	† Burr-Brown	(2850,2853)	
										3510V	† Burr-Brown	(2850,2853)	
			2	0.25	0.5	1M	110	1	Precision, Low Drift				
0.13	12	6	0.2	0.4	0.2	100K	110	0	Low Offset Voltage	OP-07D	TI		
0.15	0.075	0.01	0.7 *	1 *	0.3 *	500K	114	0	Improved LM11	LM11AC	National		
	2	0.2	2.5	0.8 *	0.12 *	80K	104	0	Improved 108 A, Low Bias, Compensated	OP-12A	† PMI	(3566)	
										OP-12E	PMI	(3566)	
								1	Improved 108 A, Low Bias	OP-08A	† PMI	(3566)	
										OP-08E	PMI	(3566)	
								110	1	Improved 108 A, Low Bias	MP0P-08A	† MicroPwr	(3529)
										MP0P-08E	MicroPwr	(3529)	
		2	0.9	0.4	0.1	300K	114	0	Instrumentation	OP-05A	† LinearTech		
				0.5 *	0.1	500K	114	0	Ultra Low Noise	OP-05A	† Raytheon	(3592)	
				0.6	0.17	300K	126	0	Low Bias, Low Drift	OP-05A	† Intersil		
				1.2 *	0.25 *	300K	114	0	Instrumentation	OP-05A	† MicroPwr	(3529)	
										MP5505A	† MicroPwr	(3529)	
										OP-05A	† PMI	(3566)	
	5	1	3	0.25 *	0.1 *	1M	94	0	Ultra Low Offset Voltage, Low Drift	AD517J	AD	(3351)	
										MP517J	MicroPwr		
	7	6	0.2	0.4	0.2	100K	100	0	Ultra-Low Offset Voltage	OP-07C	TI		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR—Operational Amplifiers—Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
0.15	7	6	1.8	0.4	0.1	120K	100	0	Low Offset Voltage, Low Drift	OP-07C	LinearTech	(Cont'd)
			0.5 *	0.1	0.1	100K	100	0	Ultra Low Noise	OP-07C	Raytheon (3592)	
			0.6 *	0.17 *		120K	100	0	Ultra Low Offset Voltage, Low Drift	ADOP-07C	AD (3351)	
				1.2 *	0.25 *	120K	100	0	Ultra Low Offset Voltage, Low Drift	OP-07C	MicroPwr (3529)	5
	7.0	6.0	0.5 *	0.5 *	0.17 *	120K	100	0	Low Noise	μPC354	NEC	
	12	6	2.5	0.5 *	0.1	100K	94	0	Ultra Low Noise	OP-07D	Raytheon (3592)	
				0.6 *	0.17 *	120K	94	0	Ultra Low Offset Voltage, Low Drift	ADOP-07D	AD (3351)	
				1.2 *	0.25 *	120K	94	0	Ultra Low Offset, Low Drift	MP5507D	MicroPwr (3529)	10
										OP-07D	PMI (3566)	
	35	20	2	0.4 *	0.5 *	1M	110	1	Low Drift	3510A	Burr-Brown (2850,2853)	
	100	4	1	0.6	0.25 *	1M	100	0	High Speed, Low Power	OP-21A	PMI (3566)	
										OP-21E	PMI (3566)	
0.17	85	90	2	5	1.7	50K	100	1	Ultra Low Noise	OP-24G	Motorola	15
0.20	7	6	0.7 *	1.2 *	0.25 *	400K	—	0	Ultra Low Offset Voltage	OP-207E	PMI (3566)	
										OP-207F	PMI (3566)	
	10	3	3	0.2 *	—	1M	114	0	Precision Instrumentation	LM363	National	
	15	15	2	0.3 *	0.1 *	1-1000	80	0	Instrumentation	AD522B	AD	
	20	5	3	1.2 *	0.25	300K	110	0	Precision, Instrumentation	CA3193A	RCA (3594)	20
	25	2.5	5	0.1 *	—	400K	100	0	Precision, Low Power	OP-20C	† PMI (3566)	
										OP-20G	PMI (3566)	
		20	6	0.3 *	0.1 *	1-1000	75	0	Instrumentation	AD522S	† AD	
	40	2	2	0.3 *	0.2	500K	90	0	Low Power	OPA21B	Burr-Brown (2850)	
										OPA21F	Burr-Brown (2850)	25
				0.5 *	49	1000K	90	1	Selectable Input	OPA201B	Burr-Brown (2850)	
										OPA201S	† Burr-Brown (2850)	
	200	20	0.5	0.6 *	0.25	1.5M	105 *	0	Precision, Low Power	OP-21G	PMI (3566)	
0.25	0.0001	0.00004	5	1 *	1.2	150K	76	0	Ultra Low Bias Military	OPA106W	† Burr-Brown (2853)	30
	0.00015	0.00004	5	0.5 *	0.3	—	80	—	Ultra Low Bias FET	3528B	Burr-Brown (2850)	
		0.00008	10	1 *	1.2	150K	76	0	Low Bias Military	OPA106V	† Burr-Brown (2853)	
	0.0003	0.00008	50	1 *	1.2	150K	76	0	Low Bias Military	OPA106U	† Burr-Brown (2853)	
	0.001	0.0002	2	1 *	0.9	100K	76	0	Low Bias, FET Input	OPA105W	† Burr-Brown (2853)	
						200K	76	0	Low Drift, Low Bias FET	OPA103DM	Burr-Brown (2850)	
			5	1 *	0.9	100K	76	0	Low Bias, FET Input	OPA105V	† Burr-Brown (2853)	35
						200K	76	0	Low Drift, Low Bias FET	OPA103CM	Burr-Brown (2850)	
			25	1 *	0.9	100K	76	1	Low Bias, FET Input	OPA105UM	† Burr-Brown (2853)	
		0.0005	3	0.7 *	0.3	40K	76	0	Precision, Low Drift FET	AD545M	AD (3351)	
			5	1 *	0.6	200K	94	0	Low Bias FET	OPA100C	Burr-Brown (2850)	
	0.002	0.0003 *	5	1 *	0.6	100K	76 *	0	Low Drift FET	3527B	Burr-Brown	40
	0.010	0.002 *	1	1.5 *	0.6	100K	90 *	0	Ultra Low Drift FET	3521L	Burr-Brown (2850)	
		0.004	5 *	10	5	50K	80	0	Low Noise, Wideband JFET	OPA101BN	Burr-Brown (2850)	
										OPA102BN	Burr-Brown (2850)	
	0.015	0.002 *	2	1.5 *	0.6	100K	90 *	0	Ultra Low Drift FET	3521K	Burr-Brown (2850)	
	0.020	0.020 *	5	1.5 *	0.6	50K	90 *	0	Ultra Low Drift FET	3521J	Burr-Brown (2850)	45
										3521R	† Burr-Brown (2850)	
	0.025	0.002	1	1	3	250K	80	2	Ultra Low Drift BIFET	AD547L	AD (3350,3351)	
	2	2	0.6	0.5 *	0.1	150K	110	0	Ultra Low Noise	OP-07A	† Raytheon (3592)	
				2.5 *	0.5	1M	100	0	Ultra Low Drift	AM-430B	Datel (2864)	
		6	0.6	0.4	0.1	300K	110	0	Ultra Low Offset, Low Drift	OP-07A	LinearTech	50
	40	35	—	63 *	11 *	—	—	0	Low Noise, High Slew Rate	OP-37E	Raytheon (3592)	
	50	0.005	5	5 *	30 *	60K	80	—	High Speed, Low Drift FET	AD381L	AD (3353)	
		35	0.5	—	5	1000	90	0	Precision Instrumentation	AD524A	AD (3354)	
	100	—	1	0.25	5	1K	100	0	Precision Instrumentation	AD624A	AD (3355)	
0.30	0.050	0.010	1 *	1 *	0.3	100K	110	0	Improved 108 A, Low Power	LM11	Motorola	55
										LM11	† National	
	0.1	0.03	5	3	10	100K	90	—	Precision JFET	MA337	AnalogSys	
	2	0.2	3.5	0.8 *	0.12 *	80K	104	0	Improved 108 A, Low Power, Compensated	OP-12B	† PMI (3566)	
										OP-12F	PMI (3566)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR—Operational Amplifiers—Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Single Units</b>										<b>(Cont'd)</b>			
0.30	2	0.2	3.5	0.8 *	0.12 *	80K	104	1	Improved 108 A, Low Power	OP-08B	† PMI (3566)		
									Improved 108 A, Low Dias	MPOP-08B	† MicroPwr (3529)		
										MPOP-08F	MicroPwr (3529)		
	5	1	1.5	—	0.7 *	1M	100	0	Programmable Micropower	OP-32E	PMI (3566)	5	
			2	—	0.7 *	1M	100	0	Programmable Micropower	OP-32A	† PMI (3566)		
	25	2	5	0.1 *	—	500K	90	0	Precision, Low Power	OP-20H	PMI (3566)		
	1000	200	2	20	10	10K	90	1	Precision, Bipolar	MA327	AnalogSys		
0.35	3	2	0.2	2.5	0.12 *	300K	104		Low Bias Current	MP5508	MicroPwr (3529)		
0.40	5	1	1.5	0.25	0.08	900K *	82	0	Programmable Micropower	OP-22A	PMI (3566)	10	
	25	20	6	0.3 *	0.1 *	1-1000	75	0	Instrumentation	AD522A	AD		
0.5	0.00075	0.00002	10	0.5 *	0.3	—	70	—	Ultra Low Bias FET	3528C	Burr-Brown		
		0.00004	10	1 *	1.6 *	100K	80	0	Ultra Low Bias FET	OPA104CM	Burr-Brown		
	0.0001	0.00005	25	1 *	0.6	100K	80 *	0	Ultra Low Bias FET	3523L	Burr-Brown	15	
	0.00015	0.00008	15	1 *	1.6 *	100K	66	0	Ultra Low Bias FET	OPA104BM	Burr-Brown		
	0.00025	0.0001	25	1 *	0.6	100K	80 *	0	Ultra Low Bias FET	3523K	Burr-Brown		
	0.0003	0.00008	10	0.5 *	0.3	—	66	—	Ultra Low Bias FET	3528A	Burr-Brown (2850)		
	0.001	0.0001	5	1 *	1.5	100K	80	0	Precision FET	LH0052	† National	20	
		0.0002	5	2	7	1M	110			1347	TeledyneP		
			5 *	0.11	4	1M	90	1	Low Bias Current, Low Power	HA-5180A-2	Harris (3452)		
										HA-5180A-5	Harris (3452)		
	0.002 *	15	1 *	0.9	200K	76	0	0	Low Drift, Low Bias FET	OPA103BM	Burr-Brown		
0.002	0.0005	10	1 *	0.6	100K	88	0	0	Low Bias FET	OPA100S	† Burr-Brown (2850)	25	
										OPA100B	Burr-Brown (2850)		
	0.003 *	25	1 *	0.9	200K	76	0	0	Low Drift, Low Bias FET	OPA103AM	Burr-Brown		
0.005	0.0003 *	2	1 *	0.6	100K	76 *	0	0	Ultra Low Drift FET	3527C	Burr-Brown (2850)	30	
		10	1 *	0.6	100K	76 *	0	0	Low Drift FET	3527A	Burr-Brown (2850)		
	0.001 *	25	1 *	0.6	50K	90 *	0	0	Low Offset FET	3522K	Burr-Brown (2850)		
										3522S	Burr-Brown (2850)		
0.01	0.01	10	—	20	800K	96	0	0	to ±150V supply, 150mA output	PA08A	Apex		
			1.3 *	3.5	100K	120 *	0	0	to ±50V supply, 5A output	PA07A	Apex		
0.015	0.006	10 *	10 *	6.5 *	50K	80	0	0	Low Noise, Wideband JFET	OPA101AM	Burr-Brown (2850)		
										OPA102AM	Burr-Brown (2850)		
0.02	0.01	10	—	100 *	—	104 *	2	0	to ±50V supply, 4A output	PA09A	Apex	35	
0.020	0.002 *	10	1.5 *	0.6	50K	90 *	0	0	Ultra Low Drift FET	3521H	Burr-Brown (2850)		
0.025	0.002	2	1	3	250K	80	2	0	Ultra Low Drift BIFET	AD547K	AD (3350,3351)		
		5	1	3	250K	80	2	0	Ultra Low Drift BIFET	AD547S	AD (3350,3351)		
			2.0	17 *	50K	80	0	0	High Speed Precision Bipolar JFET	AD544L	AD (3351)		
			5 *	30 *	40K	80	1	0	High Speed, Low Drift FET	AD382T	† AD (3353)	40	
					60K	80	—	0	High Speed, Low Drift FET	AD381T	† AD (3353)		
			10	5 *	30 *	40K	80	1	High Speed, Low Drift FET	AD382K	AD (3353)		
					60K	80	—	0	High Speed, Low Drift FET	AD381K	AD (3353)		
	0.010	5	1.0 *	3	200K	80	0	0	Precision Bipolar JFET	AD542L	AD (3351)		
	0.03	3	5 *	0.11	5	100K	100	0	Precision JFET Input	HA-5170-2	† Harris (3450)	45	
0.035	0.002	5	1.5 *	94 *	10K	80	1	0	Wideband, Fast Settling	OPA605C	Burr-Brown (2851)		
										OPA605K	Burr-Brown (2851)		
			10	1.5 *	94 *	10K	80	1	Wideband, Fast Settling	OPA605B	Burr-Brown (2851)		
										OPA605J	Burr-Brown (2851)		
0.050	0.01	5	—	45	100K	86	—	—	Wideband, Decompensated, Settles to 0.01% in 1.5 μs	OP-17A	† MicroPwr	50	
										OP-17E	MicroPwr		
										OP-17A	† PMI (3566)		
										OP-17E	PMI (3566)		
					14 *	10	100K	86	0	Bipolar-JFET, Bias Comp.	OP-15A	† MicroPwr	55
										OP-15E	MicroPwr		
										OP-15A	† PMI (3566)		
										OP-15E	PMI (3566)		
					19 *	18	100K	86	0	Wideband-JFET, Bias Comp.	OP-16A	† MicroPwr	60
										OP-16E	MicroPwr		
										OP-16A	† PMI (3566)		

† Military Temperature Range (−55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Single Units</b>										<b>(Cont'd)</b>			
0.5	0.050	0.01	5	19 *	18	100K	86	0	Wideband-JFET, Bias Comp.	OP-16E	PMI (3566)		
			5 *	0.8 *	40	60K	86	.1	Wideband-JFET, Comp. for G > 10	HA-5110-2	† Harris (3442)		
				18 *	12 *	75K	86	0	Wideband-JFET	HA-5100-2	† Harris (3440)		
										HA-5100-5	Harris (3440)		
				10 *	0.8 *	40	60K	86	0	Wideband-JFET, Comp. for G > 10	HA-5110-5	Harris (3442)	5
	0.025		10	1	1	50K	80	0	Low Power BIFET	LF441A	National		
				3	10	50K	80	0	Wideband JFET	LF411A	National		
	0.06		3	5 *	0.11	7	80K	90	Precision JFET Input	HA-5170-5	Harris (3450)		
0.1	—		50	15 *	150 *	100K	90	1	High Speed, High Power	1461-83	† TeledyneP		
0.10	0.02		5 *	2.5 *	5 *	50K	85	0	Bipolar—JFET	LF155	† AMD	10	
										LF155	† Intersil		
										LF155	† Motorola		
										PM-155	† PMI		
0.2	0.1		10 *	3 *	13 *	50K	70	0	Low Offset JFET	TL087M	† TI		
0.4	0.05		10 *	3 *	13 *	25K	70	0	Low Offset JFET	TL087C	TI	15	
	0.1		15	3 *	10	25K	—	0	Low Offset JFET	TL087	AnalogSys		
2	0.2		5	1 *	0.3 *	80K	96	1	Precision Bipolar, Low Bias	LM108A	† AMD		
										LM208A	AMD		
										AD108A	† AD (3351)		
										AD208A	AD (3351)	20	
										μA108AM	† Fairchild		
										μA208AM	Fairchild		
										LM108A	† Harris		
										LM108A	† Intersil		
										LM108A	† LinearTech	25	
										LM108A	† Motorola		
										LM208A	Motorola		
										LM108A	† National		
										LM208A	National		
										PM-108A	† PMI (3566)	30	
										PM-208A	PMI		
										SG108A	† SiliconG		
										SG208A	SiliconG		
										SFC2108A	† Thomson-CSF		
										SFC2208A	Thomson-CSF	35	
3	2.8		1	0.6	0.1	200K	114	0	Low Bias, Low Drift	OP-05	† Intersil		
			2	0.4	0.1	200K	114	0	Instrumentation	OP-05	† LinearTech		
				0.5 *	0.1	500K	114	0	Ultra Low Noise	OP-05	† Raytheon (3592)		
				0.6	0.25 *	200K	110	0	Instrumentation	MP5505B	† MicroPwr (3529)		
										OP-05	† MicroPwr (3529)	40	
										OP-05	† PMI (3566)		
4	3.8		0.9	0.5 *	0.1	500K	110	0	Ultra Low Noise	OP-05E	Raytheon (3592)		
			1.5	0.6	0.1	200K	123	0	Low Bias, Low Drift	OP-05E	Intersil		
			2	0.4	0.1	200K	110	0	Instrumentation	OP-05E	LinearTech		
				0.6	0.25 *	200K	107	0	Instrumentation	MP5505E	MicroPwr (3529)		
										QP-05E	MicroPwr (3529)	45	
										OP-05E	PMI (3566)		
7	1		5	1 *	0.3 *	80K	96	1	Precision Bipolar	LM308A	AMD		
										AD308A	AD (3351)		
										μA308AC	Fairchild	50	
										LM308A	Harris		
										LM308A	Intersil		
										LM308A	LinearTech		
										LM308A	Motorola		
										LM308A	National	55	
										PM-308A	PMI		
										SG308A	SiliconG		
										SFC2308A	Thomson-CSF		
7.5	2		2	—	0.7 *	750K	95	0	Programmable Micropower	OP-32B	† PMI (3566)		
										OP-32F	PMI (3566)	60	
30	2		8	0.8	0.25	100K	85	0	General Purpose	MP5502A	† MicroPwr (3529)		
										MP5502E	MicroPwr (3529)		
										OP-02A	† MicroPwr (3529)		
										OP-02E	MicroPwr (3529)		
			5	5	1 *	0.25 *	80K	90	1	High Accuracy 301	AD301AL	AD (3351)	65

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
0.5	35	2	8	0.8	0.25	100K	90	0	General Purpose	OP-02A OP-02E	† PMI PMI	(3566) (3566)
	40	10	5	1.2 *	0.25 *	100K	100	0	Precision, Instrumentation	CA3193 CA3493	RCA RCA	(3594) (3594)
	50	4	5	0.3 *	0.2	500K	84	0	Low Power	OPA21G	Burr-Brown	(2850)
				0.5 *	49	1000K	85	1	Selectable Input	OPA201A	Burr-Brown	(2850)
		5	5	1 *	0.5 *	50K	90	0	Higher Accuracy 741	AD741L	AD	(3351)
		30	1	1.5 *	0.8	100K *	100 *	0	Low Drift	3500E	Burr-Brown	(2850,2853)
	75	5	2	1	0.005 *	1000K	110	4	High Accuracy Instrumentation	μA725AM	† Fairchild	
	80	5	2	—	0.01	1000K	120	4	High Gain Instrumentation	OP-068 OP-06F	† PMI PMI	(3566) (3566)
		10	0.5	0.3 *	0.12 *	1000K	110	1	Precision Low Noise	AD504M	AD	(3351)
			1.0	0.3 *	0.12 *	1000K	110	1	Precision Low Noise	AD504L AD504S	AD † AD	(3351) (3351)
	500	200	15	0.3	—	—	72	0	Instrumentation	HC3020	HyComp	
0.6	0.100	0.01	5	1 *	0.3 *	100K	110	0	Precision, Low Input Current	LM11C LM11C	Motorola National	
	7.5	2	2.0	0.25	0.08	500K *	80	0	Programmable Micropower	OP-22B	PMI	(3566)
0.7	30	2	5	2.5 *	18 *	50K	90	0	High Speed	MP5501A MP5501H OP-01A OP-01H OP-01H	MicroPwr † MicroPwr † MicroPwr MicroPwr PMI	(3529) (3529) (3529) (3529) (3566)
0.75	0.075	2.8	1.3	0.5 *	0.1	150K	110	0	Ultra Low Noise	OP-07	† Raytheon	(3592)
		3.8	1.3	0.5 *	0.1	150K	106	0	Ultra Low Noise	OP-07E	Raytheon	(3592)
	4	4	1.3	2.5 *	0.5	1M	100	0	Ultra Low Drift	AM-430A AM-430M	Datal † Datal	(2864) (2864)
	80	5	—	0.44 *	0.3	50K	80	0	High Performance	μA741EC	Fairchild	
				1 *	0.3 *	50K	80	0	High Performance	μA741AM	† Fairchild	
1	0.003	0.0005	15	1 *	0.6	50K	76	0	Low Bias FET	OPA100A	Burr-Brown	(2850)
	0.01	0.01	10	—	125 *	100K	130 *	2	to ± 150V supply, 40mA output	PA84A	Apex	
				3	20	1M	130 *	0	to ± 150V supply, 75mA output	PA83A	Apex	
1.0	0.000075	—	25	0.35	0.3	50K	70	0	FET Electrometer	AD515L	AD	
	0.00015	—	15	0.35	0.3	100K	80	0	FET Electrometer	AD515K	AD	
	0.0003	0.00008	25	1 *	1.6 *	100K	66	0	Ultra Low Bias FET	OPA104AM	Burr-Brown	
	0.0005	0.0002	50	1 *	0.6	100K	80 *	0	Ultra Low Bias FET	3523J	Burr-Brown	
	0.001	0.0005	15	0.7 *	0.3	40K	70	0	Precision Low Drift FET	AD545K	AD	(3351)
	0.002	0.001	25	0.7 *	0.3	20K	66	0	Precision Low Drift FET	AD545J	AD	(3351)
	0.005	—	10	1 *	3	75K	80	0	High Accuracy, FET	AD506L	AD	
				2 *	3	50K	72	0	Low Bias Current, FET	1425-02	TeledyneP	
			25	2 *	3	50K	72	0	Low Bias Current, FET	1425-01	TeledyneP	
	0.0002	10	1 *	1	75K	76	0	Precision FET	LH0052C	National		
0.010	—	25	2 *	3	50K	72			1426-01	TeledyneP		
	0.002 *	50	1 *	0.6	50K	90 *	0	Low Offset FET	3522J	Burr-Brown	(2850)	
	0.005 *	25	10 *	65	25K	70 *	0	Fast Settling, 1 μs to 0.01%	3550J 3550S	Burr-Brown † Burr-Brown		
				20 *	100	25K	70 *	0	Fast Settling, 0.6 μs to 0.01%	3550K	Burr-Brown	
0.015	—	25	10 *	50	50K	80	0	Fast Wide Band, High Accuracy, FET Input	AD528K	AD		
								Fast Wideband, High Accuracy, FET Input	AD528S	† AD		
0.025	—	5	2 *	3	50K	72	0	Low Drift, FET	1426-03	TeledyneP		
		10	2 *	3	50K	72	0	Low Drift, FET	1426-02	TeledyneP		
	0.002	10	2	16	50K	80	0	High Speed Precision Bipolar JFET	AD544K	† AD	(3351)	
			5 *	30 *	40K	80	1	High Speed, Low Drift FET	AD382S	† AD	(3353)	
					60K	80	—	High Speed, Low Drift FET	AD381S	† AD	(3353)	
		15	2	16	50K	80	0	High Speed Precision Bipolar JFET	AD544S	AD	(3351)	
	0.01	10	1.0 *	3	50K	76	0	Precision Bipolar JFET	AD542S	† AD	(3351)	
					200K	80	0	Precision Bipolar JFET	AD542K	AD	(3351)	
0.035	0.002	25	1.5 *	94	10K	70	1	Wideband, Fast Settling	OPA605A OPA605H	Burr-Brown Burr-Brown	(2851) (2851)	

† Military Temperature Range (−55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
1.0	0.035										(Cont'd)	
		0.005	5	1	3	100K	76	2	Ultra Low Drift BIFET	AD547J	AD (3350,3351)	
	0.05	0.005	15	5 *	30 *	40K	76	1	High Speed, Low Drift FET	AD382J	AD (3353)	
						60K	76	—	High Speed, Low Drift FET	AD381J	AD (3353)	
	0.050	0.01	5 *	18 *	8 *	150K *	86 *	0	Wideband, FET	AM-410-2M	† Datal (2864)	5
				60 *	50 *	150K *	94 *	0	Wideband, FET	AM-411-2M	† Datal (2864)	
			15	70	1000	100K	44	1	150 ns Settling to 0.05%	3554B	Burr-Brown	
						30K	60	1	Wideband, Fast Settling FET	AD3554B	AD (3353)	
			25	70	1000	100K	44	1	150 ns Settling to 0.05%	3554S	Burr-Brown	
						30K	60	1	Wideband, Fast Settling FET	AD3554S	† AD (3353)	
	0.1	0.02	15	0.4	23	—	100	0	Instrumentation Amplifier, Programmable Gain 1 to 1000	AM-551MC	Datal (2861)	10
										AM-551MM	† Datal (2861)	
	0.10	0.01	50	50 *	250 *	100K	70 *	1	Wideband, Fast Settling	3551J	Burr-Brown (2851)	
										3551S	† Burr-Brown (2851)	
		0.02	—	35	75K	86	—	0	Wideband-JFET, Bias Comp.	OP-17B	† MicroPwr	15
										OP-17F	MicroPwr	
			10	—	35	75K	86	—	Wideband, Decompensated, Settles to 0.01% in 1.5 μs	OP-17B	† PMI (3566)	
										OP-17F	PMI (3566)	
				13 *	7.5	75K	86	0	Bipolar-JFET, Bias Comp.	OP-15B	† MicroPwr	20
										OP-15F	MicroPwr	
										OP-15B	† PMI (3566)	
										OP-15F	PMI (3566)	
				18 *	12	75K	86	0	Wideband-JFET, Bias Comp.	OP-16B	† MicroPwr	25
										OP-16F	MicroPwr	
										OP-16B	† PMI (3566)	
										OP-16F	PMI (3566)	
			20	40	330 *	25K	60	1	Wideband, Fast Settling	AD380K	AD (3353)	
										AD380L	† AD (3353)	
										AD380S	† AD (3353)	
	0.05	5 *	18 *	12 *	60K	80	0	0	Wideband-JFET	HA-5105-5	Harris	
	0.05 *	15 *	0.6 *	35 *	50K	80	1	1	Wideband-JFET, Comp. for G > 10	HA-5115-5	Harris	30
0.3	0.050	15	13 *	10	150K	86	0	0	Precision JFET	OP-215E	PMI (3566)	
0.30	0.02	15	40 *	220 *	25K	60	1	1	Wideband, Fast Settling	1437	TeledyneP	
										1437-80	† TeledyneP	
2	100	10	30	300 *	100K	70	2	2	Low Offset, Fast Settling	H05-060	† AD (3353)	35
										H05-060/B03	† AD (3353)	
5	0.5	10	0.8 *	0.12 *	40K	84	0	0	Precision, Low Input Current	OP-12C	† PMI (3566)	
										OP-12G	PMI (3566)	
								1	Precision, Low Input Current	MP0P-08C	† MicroPwr (3529)	40
										MP0P-08G	MicroPwr (3529)	
										OP-08C	† PMI (3566)	
										OP-08G	PMI (3566)	
10	3	3	—	0.7 *	500	85	0	0	Programmable Micropower	OP-32G	PMI (3566)	
15	7 *	3	1.5 *	1	45K	100 *	0	0	Low Bias, Low Noise	3500C	Burr-Brown (2850,2853)	
			5	1.5 *	1	45K	100 *	0	Low Bias, Low Noise	3500T	† Burr-Brown (2850,2853)	
100	20	4 *	1 *	0.25 *	30K	70	2	2	High Voltage Instrumentation	LH0004	† National	45
		5	—	—	1000K	110	4	4	Instrumentation	μPC154	NEC	
										PM-725	† PMI	
			0.5 *	0.01 *	1000K	110	4	4	Instrumentation	RM725	† Raytheon	
			1 *	0.005 *	1000K	110	4	4	High Accuracy Instrumentation	μA725M	† Fairchild	
										LM725	† National	50
200	50	6	1 *	0.4 *	25K	80	3	3	General Purpose	RM709A	† Raytheon	
250	70	20	—	—	20K	70	1	1	General Purpose	TA7502B	Toshiba	
1.2	0	3	2.0	0.35	0.08	250K *	80	0	Programmable Micropower	OP-22H	PMI (3566)	
1.3	7	6	2	0.5 *	0.1	400K	100	0	Ultra Low Noise	OP-05C	Raytheon (3592)	
			2.0	0.6	0.17	120K	100	0	Low Bias, Low Drift	OP-05C	Intersil	55
			4.5	0.4	0.1	120K	100	0	Instrumentation	OP-05C	LinearTech	
			1.2 *	0.25 *	120K	100	0	0	Instrumentation	MP5505C	MicroPwr (3529)	
										OP-05C	MicroPwr (3529)	
										OP-05C	PMI (3566)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
1.3	7	6	4.5	1.2 *	0.25 *	120K						
							120	0	Instrumentation	μPC254	NEC	(Cont'd)
		110	13	1.4 *	—	0.01	500K	110	4	High Gain Instrumentation	OP-06G	PMI (3566)
1.5	0.010	—	25	1 *	3	50K	80	0	High Accuracy, FET	AD506K	AD (3351)	
			50	1 *	3	50K	80	0	High Accuracy, FET	AD506S	† AD (3351)	
	0.10	0.05	15 *	18 *	8 *	100K *	86 *	0	Wideband, FET	AM-410-2C	Datel (2864)	5
				50 *	40 *	100K *	94 *	0	Wideband, FET	AM-411-2C	Datel (2864)	
	80	5	2	0.5 *	0.005 *	1000K	120	4	Instrumentation	LM725A	National	
	100	15	3	0.3 *	0.12 *	500K	100	1	Low Drift, Low Noise	AD504K	AD (3351)	
	120	45	4 *	1 *	0.25 *	30K	70	2	High Voltage	LH0004C	National	
2.0	0.001	0.0007	4 *	0.5 *	0.5 *	20K	55	0	Low Supply Voltage, BIMOS	CA3420B	† RCA (3594)	10
	0.001 *	0.001 *	0.7 *	0.1 *	0.04 *	30K	70	1	Programmable, Low Bias	TLC2518C	TI (3689)	
										TLC2718C	TI (541,3687,3689)	
			1 *	0.012 *	0.001 *	20K *	77	1	Prog., Low Bias, Supply Volt. = 1 V	TLC2518C	TI	
				0.075 *	0.01 *	10K *	77	1	Prog., Low Bias, Supply Volt. = 1 V	TLC2518C	TI	
			5 *	2.3 *	4.5 *	10K	70	1	Programmable, High Bias	TLC2518C	TI	15
										TLC2718C	TI	
	0.010	—	50	2 *	3	50K	72	—	Low Bias Current, FET	1425	TeledyneP	
		0.0005 *	15	1 *	3	50K	70	0	Low Offset Voltage, FET	ICL8007	† Intersil	
	0.02	—	50	100	1000	100K	90	1	Fast-Setting, FET Input	1443-83	† TeledyneP	
	0.025	—	50	2	3	50K	72	0	Low Drift, FET	1426	TeledyneP	20
	0.025 *	0.010 *	20	1.0 *	3	50K	76	0	Precision Bipolar JFET	AD542J	AD (3351)	
	0.030	0.01	4 *	0.063 *	0.03 *	32K	75	0	Manpower, BIMOS	CA3440B	RCA (3594)	
	0.05	0.01	5	2.5	3	50K	85	1	JFET Input	TDC0155A	† Thomson-CSF	
				4	10	50K	85	1	JFET Input	TDC0156A	† Thomson-CSF	
				15	40	50K	85	1	JFET Input	TDC0157A	† Thomson-CSF	25
			50	70	1000	100	60	1	Wideband, Fast Settling FET	AD3554A	AD (3353)	
		0.05	30	—	30 *	800K	96	0	to ± 150V supply, 150mA output	PA08	Apex	
				1.3 *	5	100K	120 *	0	to ± 50V supply, 5A output	PA07	Apex	
	0.050	0.005	20	2	15	30K	74	0	High Speed Precision Bipolar JFET	AD544J	AD (3351)	
		0.01	5	—	40	50K	85	—	Wideband Decompensated	LF157A	† Intersil	30
										LF357A	Intersil	
										LF157A	† Motorola	
										LF357A	Motorola	
										LF157A	† National	
										LF357A	National	35
										PM-157A	† PMI (3566)	
										PM-357A	PMI	
				2.5 *	3	50K	85	0	Bipolar - JFET	LF155A	† AMD	
										LF355A	AMD	
										LF155A	† Intersil	40
										LF355A	Intersil	
										LF155A	† Motorola	
										LF355A	Motorola	
										LF155A	† National	
										LF355A	National	45
										PM-155A	† PMI	
										PM355A	PMI	
				4	10	50K	85	0	Wideband - JFET	LF156A	† AMD	
										LF356A	AMD	
										LF156A	† Intersil	50
										LF356A	Intersil	
										LF156A	† Motorola	
										LF356A	Motorola	
										LF156A	† National	
										LF356A	National	55
										PM-156A	† PMI (3566)	
										PM-356A	PMI	
			50	70	1000	100K	44	1	150 ns Settling to 0.05%	3554A	Burr-Brown	
	0.03	10 *	1	1.6	80K *	70	0	0	Low Power	ICL7611AC	Intersil (3481)	
										ICL7611AM	† Intersil (3481)	60
									Low Power, Extended CMVR	ICL7612AC	Intersil (3481)	
										ICL7612AM	† Intersil (3481)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
2.0	0.050	0.03	10 *	1	1.6	80K *	70	0	Low Power, Input Protected	ICL7613AC	Intersil (3481)	
								1	Low Power	ICL7614AC	Intersil (3481)	
									Low Power, Input Protected	ICL7614AM	† Intersil (3481)	
									Low Power, Input Protected	ICL7615	† Intersil (3481)	
										ICL7615AC	Intersil (3481)	5
0.075	0.025	10 *	4 *	13 *	50K	80	0		Precision BIFET	MC35001A	† Motorola	
0.10	0.05	10 *	4 *	13 *	50K	80	0		Precision BIFET	MC34001A	Motorola	
									Wideband Bipolar JFET	LF351A	National	
			20	150	400	10K	60	3	Fast Settling, Wideband	OPA600N	Burr-Brown (2853)	
										OPA600V/883B	† Burr-Brown (2853)	10
										OPA600V/MIL	† Burr-Brown (2853)	
	0.05 *	40	0.5 *	3	50K	80	1		High Current, High Power	3571A	Burr-Brown	
										3572A	Burr-Brown	
	0.050	10 *	3 *	13 *	50K	80	0		Bipolar JFET	μA771A	Fairchild	
										μA771AM	† Fairchild	15
0.15	0.05	20	0.06	10	3K	100	—		JFET Input	MA403	AnalogSys	
0.2	0.02	50	40 *	225	30K	60	1		Wideband, Fast Settling	1438	TeledyneP	
										1438-83	† TeledyneP	
	0.1	20	3	10	50K	80	0		JFET Input, High Speed	TDB0351A	Thomson-CSF	
0.20	0.10	20	2.7	8	25K	70	0		Wideband JFET	LF411	National	20
		20 *	3	10	50K	80	0		Wideband Bipolar JFET	LF151A	† National	
0.30	0.02	15	40	220	25K	60	0		Wideband, settles to 0.1% in 110 ns	HC1437	HyComp	
										HC1437-883	† HyComp	
0.50	—	75	60 *	500	200K	—	0		Inverting, Settles to 0.01% in 200 ns	1430	TeledyneP	
										1430-83	† TeledyneP	25
2	0.2	15	0.3 *	0.3 *	50K	85	1		Low Noise 108	ICL108LN	† Intersil	
			1 *	0.2 *	50K	85	0		Micropower, Supply Current 600 μA	LM112	Motorola	
										LM212	Motorola	
										LM112	† National	
										LM212	National	30
					0.3 *	50K	85	1	Precision Bipolar	LM108	† AMD	
										LM208	AMD	
										AD108	† AD	
										AD208	AD (3351)	
										μA108M	† Fairchild	35
										μA208M	Fairchild	
										LM108	† Harris	
										LM108	† Intersil	
										LM108	† LinearTech	
										LM108	† Motorola	40
										LM208	Motorola	
										LM108	† National	
										LM208	National	
										PM-108	† PMI	
										PM-208	PMI	45
										SG108	† SiliconG	
										SG208	SiliconG	
										SFC2108	† Thomson-CSF	
										SFC2108M	† Thomson-CSF	
										SFC2208	Thomson-CSF	50
3	2	10	0.5 *	0.1	45K	100 *	0		Low Bias Current	3501C	Burr-Brown (2850)	
7	3	10	0.5 *	0.1	45K	100 *	0		Low Bias Current	3501B	Burr-Brown (2850)	
										3501S	† Burr-Brown (2850)	
20	0.7	2.0 *	—	—	60K	89	0		Op Amp and Voltage Reference	LM10BL	National	
					120K	93	0		Op Amp and Voltage Reference	MCELM10	MCE	55
										LM10	† National	
										LM10B	National	
	10 *	5	1.5 *	1	45K	100 *	0		Low Bias, Low Noise	3500B	Burr-Brown (2850,2853)	
		10	1.5 *	1	45K	100 *	0		Low Bias, Low Noise	3500S	† Burr-Brown (2850,2853)	
25	3	15	0.5	0.5 *	50K	80	1		Low Offset Current	μA777M	† TI	60
50	5	8	2.5 *	18 *	50K	80	0		High Speed	MP5501E	MicroPwr (3529)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Single Units</b>										<b>(Cont'd)</b>			
2.0	50	5	8	2.5 *	18 *	50K	80	0	High Speed	MP5501F OP-01E OP-01F OP-01E OP-01F	† MicroPwr MicroPwr † MicroPwr PMI † PMI	(3529) (3529) (3529) (3566) (3566)	5
			10	0.8	0.25	50K	90	0	General Purpose	MP5502 OP-02 OP-02C OP-02 OP-02C	† MicroPwr † MicroPwr MicroPwr PMI PMI	(3529) (3529) (3566)	10
			10	15	4 *	1.5	25K	80	0	High Performance, High Gain	RM4131	† Raytheon	
60	10	15		—	40	50K	70	1	60 V/μs Comp. for g=5	SE538	† Signetics		
				3 *	25	50K	70	0	High Slew Rate	SE530	† Signetics		
75	10	15		0.8 *	0.5 *	50K	80	1	Low Noise 101A	ICL101ALN	† Intersil		
				1 *	0.5 *	25K	80	1	General Purpose, Uncompensated	SFC2101A	† Thomson-CSF		
						50K	80	0	General Purpose, Compensated	LM107 LM107 LM107 LM207 LM107 LM207 LM107 SG107 SG207 LM107 LM207 SFC2107 SFC2207	† LinearTech † Motorola † Motorola † National National † Raytheon † SiliconG SiliconG † TI TI † Thomson-CSF Thomson-CSF		15 20 25
									Higher Accuracy 741	AD741S	† AD	(3351)	
								1	General Purpose, Improved 101, Uncompensated	AD101A AD201A μA101AM μA201AM LM101A LM201A LM101A LM201A LM101A SG101A SG201A LM101A LM201A SFC2201A LM101A	† AD AD † Fairchild Fairchild † Motorola Motorola † National National † Raytheon † SiliconG SiliconG † TI TI Thomson-CSF † LinearTech	(3351)	30 35 40
								90	Higher Accuracy 741C	AD741K	AD	(3351)	45
200	50	10		5 *	0.3 *	25K	80	3	General Purpose	μA709AM MC1709A LM709A SFC2709A	† Fairchild † Motorola † National † Thomson-CSF		
			25	—	0.3 *	45K *	80	3	General Purpose	μA709AM	† TI		50
800	200	—		10 *	13 *	50K	80	1	Wideband, Low Noise	XR5534M RM5534 SE5534 SE5534A SE5534	† Exar † Raytheon † Signetics † Signetics † TI	(3379) (526) (526,3646)	55
4000	1500	1.2 *		15 *	3 *	0.7K	70	1	6 Volt, Wideband	CA3010A CA3029A CA3037A	† RCA RCA † RCA		
6000	1600	1.2 *		50 *	7 *	2K	80	1	12 Volt, Wideband	CA3015A CA3030A CA3038A	† RCA RCA † RCA		60
2.5	125	35	2 *	1 *	0.005 *	250K	94	4	High Accuracy Instrumentation	μA725C LM725C μPC725	Fairchild National NEC	(Continued)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



LINEAR—Operational Amplifiers—Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
2.5	125	35	2 *	1 *	0.005 *	250K	94	4	High Accuracy Instrumentation	PM-725C RC725	PMI Raytheon	(Cont'd)
	200	40	5	0.3 *	0.12 *	250K	94	1	Low Drift, Low Noise	<b>AD504J</b>	<b>AD</b>	<b>(3351)</b>
		75	2 *	8 *	1	100K	86	0	High Performance	<b>HS3546RH</b>	<b>† Harris</b>	<b>(4714)</b>
	250	50	10 *	1 *	0.25 *	100K	90	2	High Gain Instrumentation, 50 mA	LH0020	† National	5
3	0.001	0.0002	5	2	7	1M	90			1346	TeledyneP	
										1346-01	† TeledyneP	
	0.05	0.01	10	100	120	150	74			1344-01	† TeledyneP	
			20	100	120	150	74			1344	TeledyneP	
3.0	0.0003	—	50	0.35	0.3	40K	66	0	FET Electrometer	AD515J	AD	10
	0.001	0.0002	5 *	0.11	4	1M	90	1	Low Bias Current, Low Power	<b>HA-5180-2</b>	<b>Harris</b>	<b>(3452)</b>
										<b>HA-5180-5</b>	<b>Harris</b>	<b>(3452)</b>
	0.020	0.020 *	25	3 *	20 *	100K	110 *	0	High Voltage FET	3582J	Burr-Brown	
				5	20 *	50K	110 *	0	High Voltage FET	3581J	Burr-Brown	
				5 *	20 *	400K	110 *	0	High Voltage FET	3583	Burr-Brown	15
				7 *	—	1000K	110 *	2	High Voltage FET	3584	Burr-Brown	
	0.030	—	50	10 *	50	25K	70	0	Fast Wideband, High Accuracy FET	AD528J	AD	
		0.01	10 *	5 *	13	200K	86 *	0	FET Input, Bipolar/MOS Output	<b>CA081B</b>	<b>RCA</b>	<b>(3596)</b>
								1	FET Input, Bipolar/MOS Output	<b>CA080B</b>	<b>RCA</b>	<b>(3596)</b>
	0.040	0.02	10 *	5 *	13 *	50K	80	0	FET Input, Bipolar/MOS Output	<b>CA081AT</b>	<b>† RCA</b>	<b>(3596)</b>
								1	FET Input, Bipolar/MOS Output	<b>CA080AT</b>	<b>† RCA</b>	<b>(3596)</b>
	0.05	—	75	90	900	56K	80	1	Fast-Settling, FET Input	1443	TeledyneP	
		0.01	10 *	100 *	100	75K	74	1	Wideband, High Slew Rate	<b>HA-5160-2</b>	<b>† Harris</b>	<b>(3448)</b>
			20 *	100 *	100	75K	74	1	Wideband, High Slew Rate	<b>HA-5160-5</b>	<b>Harris</b>	<b>(3448)</b>
		0.05	25	—	125 *	100K	130 *	2	to ± 150V supply, 40mA output	PA84	Apex	25
			5	20	1M	130 *	0	to ± 150V supply, 75mA output	PA83	Apex		
										PA83M	† Apex	
	0.050	0.015	—	1 *	0.3 *	40K	80	0	High Performance, Low Input Current	LM316A	National	
	0.1	0.05	30	—	100 *	—	104 *	2	to ± 50V supply, 4A output	PA09	Apex	
	0.15	0.03	10	0.11	10	3K	90	—	Low Wideband Noise	MA344	AnalogSys	30
	0.2	0.05	15	—	25	50K	82	—	Wideband, Decompensated, Settles to 0.01% in 1.6 μs	<b>OP-17C</b>	<b>† PMI</b>	<b>(3566)</b>
										<b>OP-17G</b>	<b>PMI</b>	<b>(3566)</b>
				12 *	5	50K	82	0	Bipolar-JFET, Bias Comp.	OP-15C	† MicroPwr	
										OP-15G	MicroPwr	
										<b>OP-15C</b>	<b>† PMI</b>	<b>(3566)</b>
										<b>OP-15G</b>	<b>PMI</b>	<b>(3566)</b>
				17 *	9	50K	82	0	Wideband-JFET, bias comp.	OP-16C	† MicroPwr	
										OP-16G	MicroPwr	
										<b>OP-16C</b>	<b>† PMI</b>	<b>(3566)</b>
										<b>OP-16G</b>	<b>PMI</b>	<b>(3566)</b>
				20	25	50K	82	—	Wideband, Decompensated, Settles to 0.01% in 1.6 μs	OP-17C	† MicroPwr	
										OP-17G	MicroPwr	
	0.1	—	3	13 *	50K	70	0	0	Bipolar JFET	TL081B	Motorola	
										TL081BC	TI	
			10 *	1 *	3.5 *	4K	80	0	Low Power, Bipolar-JFET	TL061BC	TI	45
									Programmable Bipolar JFET	TL066BC	TI	
				4 *	20K	80	0	0	Low Power, JFET Input	<b>HA-5062B-5</b>	<b>Harris</b>	<b>(3432)</b>
				3 *	13 *	50K	80	0	Bipolar-JFET	TL088C	TI	
										TL088M	† TI	
				4	15 *	50K	80	1	JFET Input	<b>HA-5082B-5</b>	<b>Harris</b>	<b>(3436)</b>
	0.20	0.05	10 *	4 *	13 *	50K	80	0	Low Noise Bipolar-JFET	TL071BC	Motorola	
										TL071BC	TI	
	0.30	75	10 *	0.3	10	200K *	85	0	2.0 A Power	LH0101A	† National	
										LH0101AC	National	
	0.4	0.1	15	3 *	10	25K	—	0	Low Offset JFET	TL088	AnalogSys	55
4	0.5	5	100	1000	1000K	1000K	0	—	Settles to 0.01% in 200 ns	AM-500GC	Datel	
										AM-500MC	Datel	
				7	100	1000	1000K	0	Fast Setting	<b>AM-500MB</b>	<b>Datel</b>	<b>(2861,2865)</b>
				10	100	1000	1000K	0	Fast Setting	<b>AM-500MM</b>	<b>† Datel</b>	<b>(2861,2865)</b>

† Military Temperature Range (–55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

(Continued)

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift µV/°C	Bandwidth MHz	Slew Rate V/µs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
3.0											(Cont'd)	
5	3		10 *	—	0.1 *	40K	80	0	Programmable Amplifier	HA-2720	† Harris	
7.5	3		—	0.25 *	0.16 *	100K	70	0	Programmable	ICL4250 LM4250 SG4250 UC4250	† Intersil † National † SiliconG † Solitron	5
10	2		15	0.3 *	0.13 *	50K	80	0	Micropower	RM4132	† Raytheon	
				12 *	4	100K	80	0	High Impedance	HA-2600 HA2600	† Harris † Intersil	(3406)
	1000	50	1	500	300	80	2		High Slew Rate	MA207	AnalogSys	
15	4	15	15	0.05	0.015	—	70	0	High Output, Low Power	MA112	AnalogSys	
	15	15	35 *	25	100K	80	1		Wideband, General Purpose	AD507K	AD	(3353)
20	5	—	—	0.85 *	0.55	80K	80	0	Low Power, Radiation Resistant	HS3530RH	† Harris	(4709)
	10	40	4 *	2.5	70K	74	0		to ±50V supply, 15A output	PA12A	† Apex	
			6 *	2.5	70K	74	0		to ±50V supply, 5A output	PA10A	† Apex	
25	5	10 *	—	—	4K	60	3		Matched Transistors	LH0005A	† National	15
80	30	—	—	1 *	0.6	50K	80	0	to 500 mA, single supply	µA759	† Fairchild	
		15	0.4	0.3	50K	80	0		Higher Performance	LM741A LM741E	† National National	
100	100	—	10	22	10K	80	1		High Slew Rate, Wideband	HS3516RH	† Harris	(4705)
200	50	20	1 *	0.5 *	50K	80	0		High Current	AD512K	AD	
									Higher Accuracy 741C	AD741J	AD	(3351)
									High Current	AD512S	† AD	
300	100	3 *	1 *	1.5 *	100K	70	1		0.2 A Power	LH0041	† National	
		10	1 *	0.4 *	25K	70	3		High Gain	RM709	† Raytheon	
		25	1 *	1.5	100K	70	1		1.0 A Power	LH0021	† National	25
500	75	3 *	2 *	4.2 *	50K	80	2		General Purpose	MC1539	† Motorola	
1000	200	10	40	20	100K	70	0		Low Noise, Wideband	MA322	AnalogSys	
	500	10	300	50	100K	90	2		Fast, Wideband	MA326	AnalogSys	
2000	200	4	10 *	—	20K	70	0		General Purpose Wide Bandwidth	LH0003 LH0003C	† National National	30
3.5	0.015	—	75	1 *	3	20K	70	0	High Accuracy, FET	AD506J	AD	(3351)
	12	2.5	6 *	—	0.04	150K	115 *	2	Micropower	RC3078A	† Raytheon	(3592)
4.0	0.010	0.002	10	1 *	1.5	100K	80	0	Micropower	LH0022	† National	
	3	—	12	20 *	30 *	—	70	0	Unity Gain, Noninverting	SFC2110M	Thomson-CSF	
15	2	10 *	1 *	2.5 *	100K	80	0		High Performance	MC1556	Motorola	
	15	10 *	35 *	25	100K	80	1		Wide Band, High Impedance	HA2620 HA-2620	† Intersil † Harris	
									Wideband, General Purpose	AD507S	† AD	(3353)
25	12	15 *	4 *	5 *	100K	80	0		High Voltage	HA-2640	† Harris	
30	2.0	5.0	—	—	40K	80	0		Op Amp and Voltage Reference	LM10CL	National	
					80K	90	0		Op Amp and Voltage Reference	LM10C	National	40
200	50	15	12 *	10	50K	80	0		High Speed, Fast Settling	AD518K	AD	(3353)
		20	12 *	10	50K	80	0		High Speed, Fast Settling	AD518S	† AD	(3353)
250	50	—	15 *	50	50K	80	0		Precision High Speed	LM118 LM218 LM118 LM118 LM118 LM218 LM118 LM218 SFC2118 SFC2218	† AMD AMD † Harris † LinearTech † National National † TI TI † Thomson-CSF Thomson-CSF	45
300	100	5 *	1 *	25	50K	70	1		0.5 A Wideband	LH0061	† National	
350	100	3	—	0.5 *	50K	80	0		Short Circuit Protected	TBA222	† Siemens	
				5.5 *	50K	90	1		Adjustable Input Offset Voltage	TBC0748	† Siemens	55
700 <sup>a</sup>	100	10	—	—	25K	70	1		40mA Output	MA342	AnalogSys	
		25	—	—	18K	70	1		20 mA Output	TAA762 TAA862	† Siemens † Siemens	
					9 *	18K	70	1	General Purpose, OC Output	SFC2761M SFC2861M	† Thomson-CSF † Thomson-CSF	60
1500	300	—	10 *	6 *	25K	70	1		Low noise, Comp. for G=3	XR5534	Exar	(3379)

† Military Temperature Range (−55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Single Units</b>										<b>(Cont'd)</b>			
4.0	1500	300	—	10 *	6 *	25K	70	1	Low noise, Comp. for G=3				
										XR5534A	Exar (3379)	5	
										XR5534C	Exar (3379)		
										RC5534	Raytheon (3592)		
										NE5534	Signetics		
										NE5534A	Signetics (3646)		
										NE5534	TI		
										NE5534A	TI		
						13 *	100K *	80	1	Low Noise, Comp. for G=3	AM-453-2C	Datal (2864)	
										AM-453-2M	† Datal (2864)		
	30000	5000	20 *	70 *	400	4K	60 *	3	High Slew Rate	LH0024	† National	10	
4.5	170	32	6 *	—	0.04	—	110	2	Micropower	RC3078	Raytheon (3592)	15	
					0.04 *	25K	80	1	Micropower	CA3078	† RCA		
5	15000	4000	20	150	200	30K	74			1343-01	† TeledyneP	15	
	20000	6000	20	9.5	600	30K	60			1342-01	† TeledyneP		
				400	400	30K	60			1341-01	† TeledyneP		
5.0	0.001	0.0005	5	0.7 *	0.3	40K	76	0	Precision, Low Drift FET	AD545L	AD (3351)	20	
		0.0005 *	25	1 *	0.6	50K	90 *	0	Low Offset FET	3522L	Burr-Brown		
	0.001 *	0.001 *	0.7 *	0.1 *	0.04 *	30K	70	1	Programmable, Low Bias	TLC251AC	TI (3689)	20	
										TLC271AC	TI		
				1 *	0.012 *	0.001 *	20K *	77	1	Prog., Low Bias, Supply Volt. = 1 V	TLC251AC		TI
					0.075 *	0.01 *	10K *	77	1	Prog., Low Bias, Supply Volt. = 1 V	TLC251AC		TI
				5 *	2.3 *	4.5 *	10K	70	1	Programmable, High Bias	TLC251AC		TI
										TLC271AC	TI		
	0.003	0.002	4 *	0.5 *	0.5 *	20K	70	0	Low Supply Voltage	CA3420A	† RCA (3594)	25	
	0.005	0.0006		2 *	50 *	—	80	0	Micropower, Transconductance Amplifier	CA3080	RCA		
										CA3080A	† RCA		
	0.010	0.0005 *	0.05	2 *	1 *	200K	90	0	Chopper Stabilized, Low Power	TML7650-1	Teimos	30	
					2.5 *	1M *	120	0	Chopper Stabilized	TML7650	Teimos		
	0.025	0.002	25	15 *	50	50K	80	0	Precision, High Speed FET	LH0062	† National		
	0.030	0.02	6 *	4 *	10 *	50K	80	0	MOS-Bipolar	CA3160A	† RCA (3596)	30	
			10 *	4 *	10 *	50K	80	1	MOS, Single Supply, Strobe	CA3130A	† RCA		
	0.040	0.02	4 *	0.063 *	0.03 *	10K	70	0	Nanopower BIMOS	CA3440A	RCA (3594)	35	
			6 *	4.5 *	9 *	20K	70	0	MOS FET, Single Supply	CA3140A	RCA (3595)		
	0.050	0.030	15 *	1	1.6	80K *	70	0	Low Power	ICL7611BC	Intersil (3481)	35	
										ICL7611BM	† Intersil (3481)		
										ICL7612BC	Intersil (3481)	40	
										ICL7612BM	† Intersil (3481)		
										ICL7613BC	Intersil (3481)	40	
										ICL7613BM	† Intersil (3481)		
								1	Low Power	ICL7614BC	Intersil (3481)	40	
										ICL7614BM	† Intersil (3481)		
										ICL7615BC	Intersil (3481)	45	
										ICL7615BM	† Intersil (3481)		
	0.1	0.02	5 *	2	30	25K	80	1	JFET Input	TDC0157	† Thomson-CSF	45	
				2.5	5 *	25K	80	1	JFET Input	TDC0155	† Thomson-CSF		
				5	7.5	25K	80	1	JFET Input	TDC0156	† Thomson-CSF		
	0.10	0.02	5 *	—	—	50K	85	—	Wideband Decompensated	LF157	† Intersil	50	
										LF357B	Motorola		
										LF157	† National		
										LF257	National		
										LF357B	National		
										PM-157	† PMI (3566)		
										PM-257	PMI		
				2.5 *	5 *	50K	85	0	Bipolar-JFET	LF255	AMD	55	
										LF355B	Motorola		
										LF155	† National		
										LF255	National		
										LF355B	National		
										PM-255	PMI		
				5 *	7.5 *	50K	85	0	Wideband-JFET	LF156	† AMD	60	

(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line		
<b>Single Units</b>										<b>(Cont'd)</b>				
5.0	0.10	0.02	5 *	5 *	7.5 *	50K	85	0	Wideband-JFET	LF256 LF156 LF356B LF156 LF256 LF356B <b>PM-156</b> PM-256	AMD † Intersil Motorola † National National National † PMI PMI	(Cont'd)	5	
		0.025	10 25 *	1 70 *	1 350	25K 1K	70 50	0 2	Low Power-BIFET Ultra Fast FET Ultra Fast-FET	LF441 TP0032 LH0032 ADLH0032	National † TeledyneP † National † AD		10	
		0.05	10 *	3 *	13 *	50K	80	0	Bipolar JFET Bipolar-JFET	μA771B μA771BM	Fairchild Fairchild † Fairchild		15	
				4 *	13 *	50K	80	0	Precision-BIFET	MC35001B	† Motorola			
				80	150	400	10K	60	3	Fast Settling, Wideband	<b>OPA600U</b>	<b>Burr-Brown (2853)</b>		
				100	150	400	10K	60	3	Fast Settling, Wideband	<b>OPA600U/883B</b>	<b>† Burr-Brown (2853)</b>		
0.15	—	50	20 *	210	10K	70	1	1	Wideband, Fast Settling	AH0605	OEI		20	
0.2	0.1	30	4 *	13 *	50K	80	0	0	JFET Input, High Speed	TDB0351B	Thomson-CSF		20	
0.20	0.025 0.10	3 * 10 *	1 * 4 *	0.3 * 13 *	25K 50K	96 80	0 0	0	Precision Bipolar Precision-BIFET Wideband Bipolar-JFET	LM11CL MC34001B LF351B	National Motorola National		25	
				15 *	50K	100 *	0	0	Wideband Bipolar-JFET	LF151	† National		25	
4	1	10	1 *	0.5 *	20K	80	0	0	Low Input Current 741	AD502L	AD		25	
7	4	20	1 *	0.5 *	20K	80	0	0	Low Input Current 741	AD502K	AD		25	
7.5	3	—	1 *	0.1	20K	70	0	0	Multi-Purpose, Programmable	SFC2776M	† Thomson-CSF		25	
10	5	7 * 10 *	1 * —	0.5 * 0.1 *	20K 25K	70 74	0 0	0	Low Input Current Programmable Amplifier	ICL8008M HA-2725	† Intersil Harris		30	
		20 *	1 *	0.5 *	20K	80	0	0	Darlington 741	AD502S	† AD		30	
		6	—	0.25 *	0.16 *	50K	70	0	Programmable	LM4250C ICL4250C LM4250C SG4250C UC4250C	Harris † Intersil † National † SiliconG Solitron		35	
15	5	20	0.5 *	0.1	45K	100 *	0	0	Low Bias Current	<b>3501A</b> <b>3501H</b>	<b>Burr-Brown (2850)</b> <b>† Burr-Brown (2850)</b>		35	
20	3	—	—	2.5 * 1 *	100K 100K	80 80	1 0	1	High Voltage High Voltage	LM144 MC1536 LM1536 SG1536	† National † Motorola † National † SiliconG		40	
		40	1 *	1.5	60K	80	0	0	Output 26V @ 10A	<b>OPA501B</b> <b>OPA501S</b>	<b>Burr-Brown (2850)</b> <b>† Burr-Brown (2850)</b>		40	
25	5 25	20 5 * 10 *	0.3 * 12 * 12 *	0.13 * 4 4	50K 300K 80K	70 100 74	0 0 0	0	Micropower Wide Temperature Range High Impedance	RC4132 OPA11HT <b>AM-460-2</b> <b>AM-460-2M</b> HA-2602 HA-2605 HA2602 HA2605	Raytheon † Burr-Brown (2850) Datel (2864) † Datel (2864) † Harris Harris † Intersil Intersil		45	
				15 *	35 *	20	80K	74	1	Wide Band, High Impedance	<b>AM-462-1</b> <b>AM-462-1M</b> <b>AM-462-2</b> <b>AM-462-2M</b>	<b>Datel (2864)</b> <b>† Datel (2864)</b> <b>Datel (2864)</b> <b>† Datel (2864)</b>		55
										HA2622 HA2625 HA-2622 HA-2625	† Intersil Intersil † Harris Harris		50	
		30 *	1 70	— 20	80K 80K	74 100 *	1 1	1	Wideband Wideband, High Gain	<b>3508J</b> 1321 1321-01	<b>Burr-Brown (2851)</b> TeledyneP † TeledyneP		60	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line		
<b>Single Units</b>										<b>(Cont'd)</b>				
5.0	30	15 *	20	1.5 *	0.6	45K	100 *	0	Low Bias, Low Noise	3500A	Burr-Brown (2850.2853)			
										3500R	† Burr-Brown (2850.2853)			
		30	20	2	0.4	50K	80	0	General Purpose	3500R/MIL	† Burr-Brown (2850.2853)			
			60	2	0.4	50K	80	0	General Purpose	3500U/883B	† Burr-Brown (2850.2853)			
	35	7	—	5 *	5 *	10K	67	0	High Power, Comp. G>30	CA3105	RCA	5		
										CA3105M	RCA			
	40	2.5 *	30 *	1 *	10	200K	80	0	Micropower FET High Current Out	1323	TeledyneP			
		15	5	—	20	300K	106	0	High Speed	AM-470-2C	Datel (2865)			
										AM-470-2M	† Datel (2865)			
	50	15	—	0.2	0.35	50K	70	0	Programmable Low Power	MC1776	† Motorola	10		
	100	20	10	2.5 *	18 *	25K	80	0	High Speed	MP5501C	MicroPwr (3529)			
										MP5501B	† MicroPwr (3529)			
										OP-01C	MicroPwr (3529)			
										OP-01B	† MicroPwr (3529)			
										OP-01C	PMI (3566)	15		
										OP-01B	† PMI (3566)			
			30	1 *	0.5 *	25K	70	1	Precision	μA777C	Intersil			
			25	20	1.3	0.25	25K	70	0	General Purpose	MP5502B	† MicroPwr (3529)		
										MP5502D	MicroPwr (3529)			
										OP-02B	† MicroPwr (3529)	20		
										OP-02D	MicroPwr (3529)			
										OP-02B	† PMI (3566)			
										OP-02D	PMI (3566)			
										XR146	† Exar			
	120	40	6 *	3 *	20	50K	70	0	High Slew Rate	NE530	Signetics	25		
	150	20	20	4 *	2 *	25K	70	0	High Performance, High Gain	RC4131	Raytheon			
		30	—	—	—	50K	70	0	Single Supply	TL321I	TI			
										TL321M	† TI			
		40	6 *	—	60 *	50K	70	1	60 V/μs, Comp. for G=5	NE538	Signetics			
	200	—	10 *	10 *	130	25K	—	2	High Speed Inverting	ICL8017M	† Intersil	30		
		25	20 *	12 *	25	20K	80	0	High Slew Rate	HA-2500	† Harris (3396)			
										HA2500	† Intersil			
		50	20 *	8 *	15 *	50K	80	1	Four Addressable Inputs, Single Amplifier	HA-2400	† Harris			
										HA-2404	Harris			
	250	20	20	1 *	0.5 *	25K	96	0	General Purpose	LM307	Raytheon	35		
								1	General Purpose	LM301A	Raytheon			
	500	150	20	—	—	20K	70	1	General Purpose	TA7502	Toshiba			
										TA7502A	Toshiba			
		200	—	0.5 *	35	50K	70	1	High Slew Rate	RM4531	† Raytheon			
				1 *	0.5	50K	70	0	High Performance	SFC2741M	† Thomson-CSF	40		
								1	General Purpose	SFC2748M	† Thomson-CSF			
						0.7	50K	70	0	High Speed 741	ICL741MHS	† Intersil		
						30 *	50K	70	1	High Slew Rate	SE531	† Signetics		
		3 *	0.5 *	0.3 *	25K	70	3	3	General Purpose	μA709M	† Fairchild	45		
										MC1709	† Motorola			
										LM709	† National			
						0.8 *	5 *	50K	70	0	Wideband 741	SG1217	† SiliconG	
						1 *	0.5 *	50K	70	0	General Purpose Compensated	AD741	† AD	50
										μA741M	† Fairchild			
										ICL741	† Intersil			
										MC1741	† Motorola			
										LM741	† National			
										PM-741	PMI			
										RM741	† Raytheon	55		
										μA741	† Signetics			
										SG741	† SiliconG			
										μA741M	† TI			
									Low Noise 741	ICL741LN	† Intersil			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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LINEAR—Operational Amplifiers—Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
5.0	500	200	3 *	1 *	0.5 *	50K	70	0	Low Noise 741	MC1741N CA6741	† Motorola † RCA	
								1	General Purpose, Uncompensated	μA101 LM101 LM101 SG101	† Fairchild † National † Signetics † SiliconG	5
									Uncompensated 741	μA748M MC1748 LM748 LM748C CA748 SG748 μA748M	† Fairchild † Motorola † National National † RCA † SiliconG † TI	10
					10	50K	70	0	High Slew Rate 741	MC1741S SG741S	† Motorola † SiliconG	15
									Low Noise 741	MC1741NS	Motorola	
			5 *	0.3 *	12K	80	3		General Purpose	SF2709M	† Thomson-CSF	
			4 *	30 *	0.7 *	20K	70	2	Programmable	CA3094A CA3094B	† RCA † RCA	
			6 *	—	0.3 *	45K *	70	3	General Purpose	μA709M	† TI	20
700	100	25	—	—	50 *	10K	65	0	20 mA Output	TCA322	† Siemens	
750	250	6 *	65 *	15	15K	74	3		High Speed High Gain	μA715M	† Fairchild	
1000	300	20	30	15	50K	90	0		Precision, Audio	MA332	AnalogSys	
	300 *	50	150	60	10K	70	1		High Power	1460	TeledyneP	
1000 *	5000 *	50	—	—	4000	50K	60	0	70 MHz -3db bandwidth, G=1-40 70 MHz -3db bandwidth, G=1-40, settles to 0.02% in 30 ns.	CLC210AI CLC210AM	Comlinear † Comlinear	25
					8000	50K	60	0	200 MHz -3db bandwidth, G=1-40 200 MHz -3db bandwidth, G=1-40; settles to 0.02% in 15 ns.	CLC220AI CLC220AM	Comlinear † Comlinear	
2000	400	—	38 *	25 *	0.6K	76	1		Large Signal Wideband	CA3100	† RCA (3596)	
5000 *	1000 *	20	150 *	160	15K	74	1		Wideband, Fast Settling	HA-5190	Harris	30
6000 *	3000 *	50	—	—	4000	50K	60	0	100 MHz -3db bandwidth, G=1-40 100 MHz -3db bandwidth, G=1-40; settles to 0.02% in 25 ns.	CLC200AI CLC200AM	Comlinear † Comlinear	
7000	2000	5 *	—	—	2K	70	1		702 Type	UPC51A	NEC	
10000	2000	10 *	—	—	1.7 *	1.4K	70	3	General Purpose	TL702M	† TI	
12000	5000	1.2 *	15 *	3 *	0.7K	70	1		6 Volt, Wideband	CA3010 CA3029 CA3037	† RCA RCA † RCA	35
15000	4000	10 *	150 *	200 *	28K *	—	1		Wideband, Comp. for G>5	HA-5195	Harris	
20000	—	10	—	—	800	250	70	1	Fast, Slews 800 V/μs	NE5539 SE5539	Signetics † Signetics	40
	300 *	25	500 *	250	10K	80	1		Settles to 0.01% in 70 ns	1435	† TeledyneP	
	6000	20	400 *	350	15K	60	—		Wideband, Fast Settling	HA-2540-2	† Harris (3404)	
	20 *	20 *	600 *	550	15K	60	—		High Slew Rate, Wideband	HA-2539-2	† Harris (3402)	
24000	5000	3.5 *	50 *	7 *	2K	80	1		12 Volt, Wideband	CA3015 CA3030 CA3038	† RCA RCA † RCA	45
6	15000	4000	20	150	200	30K	74			1343	TeledyneP	
6.0	0.025	0.002	15	1 *	1	75K	70	0	High Performance FET	LH0022C	National	
	0.04	0.02	10 *	5 *	13 *	50K	80	1	MOS/FET Input, Bipolar/MOS Output	CA080A CA080T CA081A CA081T	† RCA † RCA † RCA † RCA	(3596) (3596) (3596) (3596)
	0.2	—	10 *	3 *	13 *	15K	80	0	Low Noise Bipolar-JFET	TL075M	† TI	
		0.05	10 *	3 *	13	50K	80	0	Low Noise Bipolar-JFET	TL070M TL071M	† TI † TI	55
		0.1	—	3	13 *	50K	70	0	Bipolar JFET	TL081A TL081AC	Motorola TI	
			10 *	1 *	3.5 *	40K	80	0	Low Power Bipolar JFET	TL061AC TL061M	TI † TI	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Single Units</b>										<b>(Cont'd)</b>			
6.0	0.2	0.1	10 *	1 *	3.5 *	40K	80	0	Programmable Bipolar JFET	MCE66 TL066M	MCE † TI	(Cont'd)	
								1	Low Power Bipolar JFET	MCE60 TL060AC TL060M	MCE TI † TI	5	
				4 *	20K	80	0		Low Power, JFET Input	HA-5062-2 HA-5062A-5	† Harris Harris	(3432) (3432)	
				3 *	13 *	50K	80	0	Bipolar-JFET	MCE80 TL080AC	MCE TI	10	
								1	Bipolar-JFET	TL080AC	TI		
				4	15 *	25K 50K	80 80	1 1	JFET Input JFET Input	HA-5082-2 NA-5082A-5	† Harris Harris	(3436) (3436)	
				4 *	13 *	25K	80	0	JFET Input	TL081M	† Motorola		
0.20	0.05	10 *	4 *	13 *	50K	80	0		Low Noise Bipolar-JFET	TL071AC TL071AC	Motorola TI	15	
0.3	0.1	10 *	1 *	3.5 *	4K	80	0		Programmable Bipolar JFET	TL066AC	TI		
10	6	—	—	1 *	0.1	50K	70	0	Multi-Purpose, Programmable	SFC2776C	Thomson-CSF		
25	10	6 *	—	—	0.2 *	75K	70	0	Programmable	SG3250	SiliconG		
	12	40	1 *	—	0.5 *	20K	80	0	Low Input Current 741	AD502J	AD		
	20	15 *	—	—	0.5 *	20K	70	0	Low Input Current	ICL8008C	Intersil	20	
	25	15 *	35 *	20	80K	74	1		Wideband, General Purpose	AD507J	AD	(3353)	
30	30	—	—	0.05	4	70K	74	0	Wideband	HA2607	Intersil		
		15 *	—	0.6	17	70K	74	0	Wideband	HA2627	Intersil		
				4 *	5 *	100K	74	0	High Voltage	AM-464-2 AM-464-2M HA-2645 1332	Datel † Datel Harris TeledyneP	25	
				65	4 *	2.5	70K	74	0	to ±45V supply, 5A output	PA12 PA12M	Apex † Apex	30
					6 *	2.5	70K	74	0	to ±45V supply, 5A output	PA10 PA10M	Apex † Apex	
75	10	—	—	0.4	1.5	100K	80	—	Ultra-Low Power	HA-5141A	Harris	(3446)	
	20	—	—	—	—	60K	70	0	Programmable	μPC4250C	NEC		
100	30	5 *	—	—	0.3 *	25K	70	1	Supply to ±10 V	TCA520B	Signetics		
200	50	10	1	50	32	80	1		Differential Bipolar Input	MA318	AnalogSys	35	
250	50	—	1 *	0.5	25K	80	0		to 500 mA, Single Supply	μA759C	Fairchild		
500	200	—	—	0.5 *	5K	70	0		Short Circuit Protected	TBA221	Siemens		
				5.5 *	5K	70	1		Adjustable Input Offset Voltage	TBB0748	Siemens		
				1 *	0.5	20K	70	0	High Performance	SFC2741C	Thomson-CSF		
								1	General Purpose	SFC2748C	Thomson-CSF	40	
						20-50K	70	0	General Purpose Compensated	AD741C IC450 μA741C HA17741 MC1741C LM741C μPC151 μPC741 PM-741C RC741 CA741C μA741C SG741C μA741C	AD Cherry Fairchild Hitachi Motorola National NEC NEC PMI Raytheon RCA Signetics SiliconG TI	(3351)	
				0.5 *	20-50K	70	1		General Purpose	μA748C μA748C MC1748C CA748C SG748C μA748C	Fairchild Intersil Motorola RCA SiliconG TI	55	
						25K	70	0	High Speed 741	ICL741CHS	Intersil	60	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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Master Selection Guide

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
6.0	500	200		1*	1	100K	70	1	0.2 amp Power	LH0041C	National	(Cont'd)
				3*	1*	10	20K	70	0	High Slew Rate 741	MC1741SC SG741SC	Motorola SiliconG
										Low Noise 741	MC1741NC	Motorola
				5*	0.8*	5*	20K	70	0	Wideband 741C	SG3217	SiliconG
				10*	1*	0.25	100K	90	2	High Gain, Instrumentation, 50 mA Output	LH0020C	National
						0.5*	20K	—	0	Automotive Temperature Range (-40°C to 85°C)	SA741C	Signetics
							25K	70	0	Low Noise 741	ICL741CLN	Intersil
				15*	0.2*	0.2*	20K	70	0	High Power	μA791C TBA0791	Fairchild Thomson-CSF
				30	1*	1*	100K	70	0	1.0 amp Power	LH0021C	National
	300	10*	12*	50	25K	70	0	0	High Speed, Fast Setting	AD518J	AD	(3353)
1000	200	6	1*	9*	10K	70	1	1	General Purpose	SFC2861AM	† Thomson-CSF	
	300	6*	—	9*	12K	65	1	1	General Purpose, OC Output	SFC2761C	Thomson-CSF	
				18*	12K	65	1	1	20 mA Output	TAA761	Siemens	
1200	500	5*	4*	100*	10K	70	2	2	Gated	ZN424	Ferranti	
1500	200	—	0.5*	35*	20K	70	1	1	High Slew Rate	RC4531	Raytheon	
			1*	20*	20K	70	1	1	High Slew Rate, High Performance	NE531	Signetics	
7.0	100	10	—	0.4	1	100K	77	—	Ultra-Low Power	HA-5141	Harris	(3446)
	200	10*	10*	10*	130*	25K	—	2	High Speed Inverting	ICL8017C	Intersil	
	400	75	5*	—	—	3K	63	0	250 mA Output, Electronic Shutdown	LM13080	National	
				1*	0.4*	3K	63	0	250 mA Output, Electronic Shutdown	LM13080	AnalogSys	
	6000	75	5*	—	1.6	3K	63	0	450 mA Output, Electronic Shutdown	MA324	AnalogSys	
7.5	7	—	10*	20*	30*	—	70	0	Unity Gain, Noninverting	SFC2310	Thomson-CSF	
		1	—	0.3*	0.3*	25K	80	1	Low Noise 308	LM308LN	Intersil	
			30	1*	0.3*	25K	80	0	Micropower, Supply Current 800 μA	LM312	Motorola	
										LM312	National	
								1	Precision Bipolar	LM308	AMD	
										AD308	AD	(3351)
										LM308	Harris	
										LM308	Intersil	
										LM308	LinearTech	
										LM308	Motorola	
										LM308	National	
										PM-308	PMI	
										SG308	SiliconG	
										SFC2308	Thomson-CSF	
	250	50	—	0.8*	0.5*	25K	70	1	Low Noise 301A	ICL301ALN	Intersil	
			30	1*	0.5*	25K	70	0	General Purpose, Compensated	LM307	Intersil	
										LM307	LinearTech	
										LM307	Motorola	
										LM307	National	
										CA307	RCA	
										SG307	SiliconG	
										LM307	TI	
										SFC2307	Thomson-CSF	
								1	General Purpose Uncompensated	AD301A	AD	(3351)
										μA301A	Fairchild	
										LM301A	LinearTech	
										LM301A	Motorola	
										LM301A	National	
										μPC157	NEC	
										μPC301	NEC	
										CA301A	RCA	
										SG301A	SiliconG	
										LM301A	TI	
										SFC2301A	Thomson-CSF	
1000	150	3*	2*	4.2*	15K	80	2	2	General Purpose	MC1439	Motorola	
	300	6*	—	18*	5K	60	0	0	20 mA Output	TCA321	Siemens	
										TCA325	Siemens	
1500	250	6	65*	10	10K	74	3	3	High Speed	μA715C	Fairchild	
	500	6*	—	0.3*	15K	65	3	3	General Purpose	μA709C	TI	
			1*	0.5*	20K	65	1	1	General Purpose	μA201	Fairchild	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.





LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line									
<b>Single Units</b>										<b>(Cont'd)</b>											
10.0	0.20	0.1	10 *	4 *	13 *	25K	70	0	Precision BIFET	MC34001	Motorola	(Cont'd)									
										TL075C	TI										
										LF400	National										
0.60	0.1	20 *	20 *	18 *	57 *	200K	80	0	Wideband-JFET	LH0101	National	5									
1	0.25	10	10	5	10	50K	85	0	2.0 A Power	LH0101C	National										
										9906	† OEI										
10	3	150	—	—	250	0.7K	50	1	100 MHz	MC1456	Motorola										
30	10	—	—	1 *	2.5 *	70K	70	0	High Performance	MC1436	Motorola										
40	10	—	—	1 *	2 *	70K	70	0	High Voltage	SG1436	SiliconG										
										3573	Burr-Brown	10									
										OPA501R	† Burr-Brown (2850)										
										OPA501A	Burr-Brown (2850)										
										500K	70	0									
										PA11	Apex										
										PA73	Apex										
50	20	20 *	—	—	—	2K	55	3	Matched Input Transistors	LH0005	National	15									
100	25	25 *	—	—	—	2K	50	3	Matched Input Transistors	LH0005C	National										
250	50	20 *	20 *	20 *	80	7.5K	74	1	High Speed, Fast Settling	AD509J	AD										
		25 *	12 *	15	15K	74	0	0	High Slew Rate	HA2507	Intersil										
					40	7.5K	74	0	High Slew Rate	HA-2512	† Harris										
										HA2512	Intersil	20									
										HA-2522	† Harris										
										HA2522	† Intersil										
										3507J	Burr-Brown										
										HA2517	Intersil										
										HA-2515	Harris	25									
										HA2515	Intersil										
										HA-2525	Harris										
										HA2525	Intersil										
										80 V/μs Gain=3											
										AM-452-2	Datel (2864)	30									
										AM-452-2M	† Datel (2864)										
										20 *	60	7.5K	74	0	High Slew Rate	HA2527	Intersil				
										80	7.5K	90 *	1	1	High Slew Rate	1322	TeledyneP				
															1322-01	† TeledyneP					
															70K	90 *	0	High Slew Rate/Temperature	OPA12HT	† Burr-Brown (2851)	
500	200	—	—	15 *	50	25K	70	0	Precision, High Speed	LM318	AMD	35									
										μA318	Fairchild										
										LM318	Harris										
										LM318	LinearTech										
										LM318	National										
										LM318	TI	40									
										SFC2318	Thomson-CSF										
										5 *	1 *	25	25K	60	1	0.5 A, Wideband	LH0061C	National			
										300	10	2	10	50K	70	1	Up to ±40 V Output	MA700	AnalogSys		
750	—	—	—	—	—	14K	—	1	Telephone Channel Amplifier	LS045	SGS	45									
1000	300	6 *	—	—	9 *	5.6K	60	1	General Purpose, OC Output	SFC2861C	Thomson-CSF										
						18 *	0.56K	60	1	to 90 mA Output	TAA861	Siemens									
										TAA865	Siemens										
1000 *	5000 *	50	—	—	6000	5K	60	0	150 MHz -3db bandwidth, G=1-40	CLC103AI	Comlinear										
1200	300	6	1 *	—	9 *	5.6K	65	1	General Purpose	SFC2861AC	Thomson-CSF										
3000	300	16	20	10	100K	60	2	Up to 400 mA Output	MA206	AnalogSys	50										
15000	5000	5 *	—	—	1.7 *	1K	65	3	General Purpose	TL702C	TI										
12.0	50	15	65	1 *	1.5	500K	70	0	to ±28V supply, 5A adjustable output	PA01	Apex										
	90	25	—	1 *	2 *	50K	50	0	High Voltage	MC1436C	Motorola										
										SG1436C	SiliconG										
										30	—	1 *	2.5	25K	110 *	0	High Performance	MC1456C	Motorola	55	
										LM1436	National										
14.0	200	15	50 *	—	—	10K	65	1	Darlington Input	TCA312	† Siemens										
										TCA332	† Siemens										
15	0.065	0.01	35	100	70	100	70			1345	TeledyneP										
	20000	6000	20	9.5	600	30K	60			1342	TeledyneP										
				400	400	30K	60			1341	TeledyneP										

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
15	35000	—	33	80	1300	1				1359 1359-01	TeledyneP † TeledyneP	(Cont'd)
15.0	0.010	—	25	2 *	3	50K	72	0	High Performance, Low Bias	1421-02	TeledyneP	
	0.015	—	50	2 *	3	50K	72	0	High Performance, Low Bias	1421-01	TeledyneP	
	0.050	—	50	2 *	3	50K	72	0	High Performance FET	1421	TeledyneP	5
	0.030	—	8 *	4 *	10 *	50K	70	0	MOS FET, Single Supply	CA3160	† RCA	
				4.5 *	9 *	20K	70	0	MOS FET, Single Supply	CA3140	† RCA	
			10 *	4 *	10 *	50K	70	1	MOS FET Single Supply, Strobe	CA3130	† RCA	
				5 *	13 *	50K	70	0	FET Input, Bipolar/MOS Output	CA081	RCA (3596)	10
									FET Input, Bipolar/MOS Output	CA081C	RCA (3596)	
								1	FET Input, Bipolar/MOS Output	CA080	RCA (3596)	
										CA080C	RCA (3596)	
			25 *	1	1.6	80K	70	0	Low Power	ICL7611C	Intersil	
									Low Power, Extended CMVR	ICL7612C	Intersil	
									Low Power, Input Protected	ICL7613DC	Intersil (3481)	15
								1	Low Power	ICL7614DC	Intersil (3481)	
									Low Power, Input Protected	ICL7615DC	Intersil (3481)	
0.065	0.005	—	35	15 *	50	25K	70	0	Precision, High Speed, FET	LH0062C	National	
	0.01	—	20 *	100 *	50	25K	70	1	Wideband, High Slew Rate	HA-5162-5	Harris	
0.20	0.05	—	10 *	1 *	0.5 *	25K	70	0	JFET-741	LF13741	National	20
			25 *	50 *	350	1K	50	2	Ultra Fast FET	LH0032C	National	
										ADLH0032C	AD	
	0.10	—	10 *	3 *	13 *	50K	70	0	Bipolar-JFET	μA771L	Fairchild	
0.4	0.05	—	—	3 *	13 *	25K	70	1	Low Noise Bipolar-JFET	MCE70	MCE	
	0.2	—	—	3	13 *	25K	70	0	Bipolar JFET	TL081	Motorola	25
										TL081	TI	
			10 *	1 *	3.5 *	3K	70	0	Low Power Bipolar-JFET	TL061C	TI	
									Programmable Bipolar JFET	TL066C	TI	
								1	Low Power Bipolar-JFET	TL060C	TI	
				4 *	10K	70	0	0	Low Power, JFET Input	HA-5082-5	Harris (3432)	30
				3 *	13 *	25K	70	0	Bipolar-JFET	μPC4081	NEC	
								1	Bipolar-JFET	TL080C	TI	
				4	15 *	25K	70	—	JFET Input	HA-5082-5	Harris (3436)	
50	25	—	12 *	—	—	2K	60	1	High Output Current	SFC2315	Thomson-CSF	
750	150	—	30	300	15	10K	84	1	High Power	MA329	AnalogSys	35
20000	6000	—	20	400 *	350	10K	60	—	Wideband, Fast Settling	HA-2540-5	Harris (3404)	
			20 *	600 *	550	10K	60	—	High Slew Rate, Wideband	HA-2539-5	Harris (3402)	
20.0	0.00025	—	60	0.5 *	3	40K	80	0	Electrometer, FET	AD523L	AD	
	0.0005	—	30	0.5 *	3	40K	80	0	Electrometer, FET	AD523K	AD	
	0.010	—	25	0.75 *	0.5 *	50K	70	0	Low Noise FET	AD514L	AD	40
	0.002 *	—	25	1 *	2.5	31K	86 *	0	High Performance, Low Bias FET	3503B	Burr-Brown	
										3503S	† Burr-Brown	
	0.020	—	25	0.75 *	0.5	50K *	70	0	Low Noise FET	AD514K	AD	
			50	0.75 *	0.5 *	50K *	70	0	Low Noise FET	AD514S	† AD	
	0.0005 *	—	75	1 *	6 *	50K	70	0	High Performance FET	ICL8007M	† Intersil	45
0.025	—	—	25	1 *	6 *	50K	70	0	Low Cost FET Input	AD540K	AD (3351)	
			50	1 *	0.5 *	25K	80	0	FET	AD3542J	AD	
					6 *	50K	70	0	Low Cost FET Input	AD540S	† AD (3351)	
	0.002 *	—	50	1 *	0.5 *	25K	80 *	0	Low Noise	3542J	Burr-Brown (2850)	
										3542S	† Burr-Brown (2850)	50
	0.010	—	20	1 *	1.5	50K	70	0	Low Cost FET Input	LH0042	† Intersil	
										LH0042	† National	
0.050	—	—	25	1 *	6 *	25K	70	0	Low Cost FET	AD0042C	AD	
	0.005	—	25	1 *	1	25K	70	0	Low Cost FET	LH0042C	Intersil	
										LH0042C	National	55
0.10	—	—	25	1 *	3	50K	80	0	Low Cost FET Input	AD503K	AD (3351)	
										AD503S	† AD (3351)	
0.20	0.15	—	20 *	3 *	6 *	25K	64	0	High Slew Rate FET	μA740	† Intersil	
0.30	—	—	100	10 *	250 *	80	80		FET Input	AH0008	OEI	

(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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Master Selection Guide

LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Single Units</b>										<b>(Cont'd)</b>		
20.0	50	20	12 *	—	9 *	0.5K	60	1	Darlington Input	TBB1331A	Siemens	(Cont'd)
	200	25	12 *	—	—	6K	60	0	Darlington Input	TCA311 TCA315	Siemens Siemens	
								1	Darlington Input	TCA331	Siemens	
30.0	0.001	0.0002 *	50	1 *	2.5	20K	86	0	Low Bias FET	ICL8007AC ICL8007AM	Intersil † Intersil	5
	30000	10000	150	200	300	1K	50	0	Wideband	9916	† OEI	
	50000	20000	100	300	1K	4K	90	1	Wideband	9914	OEI	
35.0	2	0.1	150	100	300	100K	70	0	High Slew Rate	HDS-060	AD	
50.0	0.00001	—	—	—	0.5 *	20K	60	0	Ultra low Bias FET	ICH8500A	Intersil	10
	0.0001	—	—	—	0.5 *	20K	60	0	Ultra low Bias FET	ICH8500	Intersil	
	0.001	0.0005 *	90	0.5 *	3	20K	70	0	Electrometer, FET	AD523J	AD	
	0.015	—	75	1 *	3	20K	70	0	Low Cost FET Input	AD503J	AD	(3351)
	0.025	0.005 *	75	1 *	2.5	20K	86 *	0	High Performance, Low Bias FET	3503A 3503R	Burr-Brown † Burr-Brown	15
	0.050	—	75	0.75 *	0.5 *	20K	70	0	Low Noise FET	AD514J	AD	
				1 *	6 *	20K	70	0	Low Cost FET	AD540J	AD	(3351)
							72	0	High Performance FET	1424	TeledyneP	
		0.0005 *	75	1 *	6 *	20K	70	0	High Performance, FET	ICL8007C	Intersil	
110.0	0.20	0.15	20 *	3 *	6 *	25K	64	0	High Slew Rate FET	μA740C	† Intersil	20
<b>Dual Units</b>												
—	8000 *	—	—	30	60 *	4K	—	0	GB Product = 400 MHz	LM159 LM359	† National National	
0.06	3	2.8	0.9	0.4	0.1	400K	110	0	Ultra Low Offset Matched	LT1002AM LT1002AC	† LinearTech LinearTech	
0.08	40	35	1	5	1.7	250K	114	0	Ultra Low Noise Instrumentation	MPOP-227A MPOP-227E OP-227A OP-227E	† MicroPwr MicroPwr † PMI PMI	(3529) (3529) (3566) (3566)
0.1	3.5	3.5	0.4	1.2	0.25	500K	—	0	Ultra Low Offset Voltage	MPOP-207A MPOP-207E	† MicroPwr MicroPwr	(3529) (3529)
0.10	4.5	4.2	1.3	0.4	0.1	350K	110	0	Ultra Low Offset Matched	LT1002C LT1002M	LinearTech † LinearTech	
	5	0.5	10	0.8	1.12 *	40K	104	0	Precision, Low Input Current	MPOP-12C MPOP-12G	† MicroPwr MicroPwr	(3529) (3529)
0.12	55	50	1.5	5	1.7	250K	106	0	Ultra Low Noise Instrumentation	MPOP-227B MPOP-227F OP-227B OP-227F	† MicroPwr MicroPwr † PMI PMI	(3529) (3529) (3566) (3566)
0.15	2	0.2	2.5	0.8	1.12 *	80K	104	0	Improved 108A, Low Bias, Compensated	MPOP-12A MPOP-12E	† MicroPwr MicroPwr	(3529) (3529)
	7	6	1.8	1.2 *	0.25 *	120K	100	0	Ultra Low Offset Voltage, Low Drift	μA714C MP5507C OP-07C	Fairchild MicroPwr PMI	(3529) (3529) (3566)
	7.0	6.0	0.5 *	0.5 *	0.17 *	120K	100	0	Low Noise	μPC45A	NEC	
	20	2	2	100 *	—	500K	96	0	Low Power, Single/Dual Supply	OP-220A OP-220E	† PMI PMI	(3566) (3566)
	80	3	1.5	—	0.2	1.5M	95	0	Low Power, Precision Matched	OP-221A OP-221E	† PMI PMI	
0.18	80	75	1.8	5	1.7	200K	106	0	Ultra Low Noise Instrumentation	MPOP-227C MPOP-227D	† MicroPwr MicroPwr	(3529) (3529)
			2	5	1.7	200K	106	0	Ultra Low Noise Instrumentation	OP-227C OP-227G	† PMI PMI	(3566) (3566)
0.25	0.035	0.002	2.5	1 *	3 *	250K	80	—	Ultra Low Drift BIFET	AD647L	AD	(3350,3351)
0.3	0.050	0.010	3	1 *	0.3 *	390K	110	0	Dual LM11, High Performance	LH2011	† National	
	6	6	1.3	1.2	0.25	400K	—	0	Ultra Low Offset Voltage	MPOP-207B MPOP-207F	† MicroPwr MicroPwr	(3529) (3529)
	25	2.5	5	100 *	—	300K	90	0	Low Power, Single/Dual Supply	OP-220B OP-220F	† PMI PMI	(3566) (3566)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR—Operational Amplifiers—Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Dual Units</b>										<b>(Cont'd)</b>		
0.3	100	5	2	—	0.2	1M	90	0	Low Power, Precision Matched	OP-221B OP-221F	† PMI PMI	(Cont'd)
0.30	2	0.2	3.5	0.8	1.12 *	80K	104	0	Improved 108A, Low Power, Compensated	MPOP-12B MPOP-12F	† MicroPwr MicroPwr	(3529) (3529)
0.5	0.035	0.002	5	1 *	3 *	250K	80	—	Ultra Low Drift BIFET	AD647K AD647S AD647S/883B	AD (3350,3351) † AD (3350,3351) † AD (3350,3351)	5
		0.005 *	5	2 *	3 *	250K	80	0	Dual 544	AD644L	AD	(3351)
			5 *	1 *	3 *	300K *	80	0	Dual 542	AD642L	AD	(3351)
			15	2 *	3 *	250K	80	0	Dual 544	AD644S	† AD	(3351)
0.06	0.025	5 *	—	13 *	50K	80	0	0	Precision BIFET	MC35022A	† Motorola	10
0.075	0.03	5 *	—	13 *	50K	80	0	0	Precision BIFET	MC34003A MC34022A	Motorola Motorola	10
2	0.2	15	1 *	0.3 *	80K	96	1	1	Dual 108A	LH2108A LH2108A LH2208A PM-2108A PM-2208A	† Intersil † National National † PMI (3566) PMI	15
3	2.8	2	0.6 *	0.17 *	150K	110	0	0	Dual Matched Instrumentation	OP-10 OP-10A OP-10E	† MicroPwr † MicroPwr MicroPwr	(3529) (3529) (3529)
7	1	30	1 *	0.3 *	80K	96	1	1	Dual 308A	LH2308A LH2308A PM-2308A	Intersil National PMI	20
20	0.1	10	3	13	50K	80	0	0	Bipolar-JFET	TL287C TL287M	TI † TI	25
120	7	3	—	0.2	800K	80	0	0	Low Power, Precision Matched	OP-221C OP-221G	† PMI PMI	25
0.6	0.10	0.01	2	0.8 *	0.3	250K	110	0	Dual LM11, High Performance	LH2011B	† National	30
40	4	10	100 *	—	500 *	76	0	0	Low Power, Single/Dual Supply	OP-220C OP-220G	† PMI PMI	(3566) (3566)
0.75	50	2	8	0.8	0.5	100K	90	0	Dual Matched	OP-04A OP-04E OP-14A OP-14E	† PMI (3566) PMI (3566) † PMI (3566) PMI (3566)	35
									OP-04 with Internally Connected Vcc Terminals	MC35022	† Motorola	35
		5	8	0.8	0.25	100K	80	0	Dual Matched	MPOP-03E MPOP-04A MPOP-04E MPOP-14A MPOP-14E	MicroPwr (3529) † MicroPwr (3529) MicroPwr (3529) † MicroPwr (3529) MicroPwr (3529)	40
									MPOP-04 With Internally Connected Vcc Terminal	MPOP-03A	† MicroPwr (3529)	40
1.0	0.035	0.005 *	10	1 *	3 *	300K *	80 *	0	Dual 542	AD642K AD642S	AD (3351) † AD (3351)	45
				2 *	3 *	250K	80	0	Dual 544	AD644K	AD	(3351)
0.05	0.025	10	1	1	50K	80	0	0	Low Power BIFET	LF422A LF442A	National National	50
0.075	0.005	10	1 *	3 *	100K	76	—	—	Ultra Low Drift BIFET	AD647J	AD (3350,3351)	50
	0.050	5 *	4 *	13 *	50K	80	0	0	BIFET	MC35022B	† Motorola	50
0.18	0.02	3	0.8 *	0.3	90K	—	0	0	Dual LM11, High Performance	LH2011C	National	50
0.2	0.1	10 *	4 *	13 *	25K	70	0	0	Bipolar-JFET	LF153 LF253	Harris Harris	50
	0.10	10	3	10	50K	80	0	0	Wideband JFET	LF412A	National	50
40	5	10	0.15 *	—	300K	86 *	0	0	Micropower, Precision	OP-220H	PMI	(3566)
500	600	—	2 *	50 *	20K	80	0	0	Transconductance Amplifier	LM11700A LM13700A	National National	55
1.3	7	6	4.5	1.2 *	0.25 *	120K 250K	100 100	0 0	Dual Matched Instrumentation Dual Matched Instrumentation	OP-10C MP5510C OP-10C	PMI MicroPwr (3529) MicroPwr (3529)	55

† Military Temperature Range (–55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Dual Units</b>										<b>(Cont'd)</b>			
2.0	0.02	0.01	5 *	4 *	10 *	50K *	86	0	Dual 3160	CA3260B	RCA (3595)	5	
	0.05	0.030	10 *	1	1.6	80K *	70	0	Low Power, Dual 747	ICL7622AC	Intersil (3481)		
									Low Power, Dual 1458	ICL7621AC	Intersil (3481)	5	
										ICL7621AM	† Intersil (3481)		
			0.1	20	3	10	50K	80	0	Dual 0351A	TDB0353A	Thomson-CSF	10
0.075	0.025	0.030	10 *	4 *	13 *	50K	80	0	Precision BIFET	MC35002A	† Motorola		
			5 *	4 *	13 *	50K	80	0	Precision BIFET	MC34022A	Motorola	10	
	0.075 *	20 *	1 *	3 *	200K *	76 *	0	0	Dual 542	AD642J	AD (3351)		
					2.0	3 *	100K	76	0	Dual 544	AD644J	AD (3351)	
0.10	0.05	10 *	3 *	13 *	50K	80	80	0	Bipolar-JFET	μA772A	Fairchild	15	
										μA772AM	† Fairchild		
			4 *	13 *	50K	80	80	0	Dual 351A	LF353A	National	15	
									Precision BIFET	MC34002A	Motorola		
0.15	0.05	20	0.06	10	3K	100	—	—	JFET Input	MA406	AnalogSys	20	
	0.07	5 *	4 *	13 *	25K	80	0	0	Precision BIFET	MC35022	† Motorola		
0.2	0.1	10 *	4 *	13 *	25K	80	80	0	Bipolar-JFET	LF353A	Harris	20	
2	0.2	15	1 *	0.3 *	50K	85	1	1	Dual 108	LH2108	† National		
										LH2208	National	20	
										PM-2108	PMI (3566)		
										PM-2208	PMI	25	
50	10	15	1 *	—	50K	70	0	0	Low Power, Single Supply	CA158A	† RCA		
	25	5	1.5 *	0.8	100K *	100 *	—	—	Matched Amplifier Pairs	3500MP	Burr-Brown	25	
										(2850.2853)			
75	5	10	0.8	0.5	50K	80	80	0	Dual Matched	MPOP-04	† MicroPwr (3529)	25	
										MPOP-04C	MicroPwr (3529)		
										MPOP-14C	MicroPwr (3529)	30	
									MPOP-04 With Internally Connected Vcc Terminal	MPOP-03	† MicroPwr (3529)		
										MPOP-03C	MicroPwr (3529)	30	
									90	0	Dual Matched		
										OP-04	PMI (3566)	30	
										OP-04C	PMI (3566)		
										OP-14C	PMI (3566)	35	
			10	15	1 *	0.5 *	50K	80	1	Dual LM101A High Performance	LH2101A	Intersil	35
										LH2101A	† National		
										LH2201A	National	35	
										LH2101A	Raytheon		
										LH2101A	Signetics (526)	40	
100	20	10	5.7 *	16 *	220K	100 *	0	0	Precision Bipolar-JFET	OP-215B	† PMI		
										OP-215F	PMI	40	
200	75	3 *	8 *	1	100K	86	86	0	Low Noise, High Performance	HA-5102-2	† Harris (3459)		
										HA-5102-5	Harris (3459)	40	
			60 *	12	100K	86	86	1	Wideband	HA-5112-2	† Harris (3459)		
										HA-5112-5	Harris (3459)	45	
800	200	—	10 *	13	50K	80	80	1	Dual 5534	SE5532	† Signetics (526.3644)		
										SE5532A	† Signetics (526.3644.3644)	45	
5000	600	7 *	2 *	50 *	—	80	80	0	Transconductance Amplifier, G = 12000 μmho.	LM13600A	Signetics (3636)		
3.0	0.03	0.01	10 *	5 *	13 *	50K	80	0	MOS/FET Input, Bipolar/MOS Output	CA082B	RCA (3596)	50	
										CA083B	RCA (3596)		
	0.04	0.02	10 *	5 *	13 *	50K	80	0	MOS/FET Input, Bipolar/MOS Output	CA082AT	† RCA (3596)	50	
	0.15	0.03	10	0.11	10	3K	90	—	Low Wideband Noise	MA345	AnalogSys		
	0.2	0.1	10 *	1 *	3.5 *	4K	80	0	Low Noise Bipolar-JFET	TL062BC	TJ	50	
				3 *	15 *	50K	80	0	Bipolar-JFET	TL082B	Motorola		
										TL082BC	TJ	55	
										TL288C	TJ		
										TL288M	† TJ	55	
			20	0.7	8	25K	70	0	Wideband JFET	LF412	National		
	0.20	0.05	10 *	4 *	13 *	50K	80	0	Low Noise Bipolar-JFET	TL072BC	Motorola	55	
5	3	10 *	25 *	0.1 *	40K	80	80	0	Programmable	HA-2730	† Harris		
	5	—	1	1	100K	100	100	0	Low Noise	SE5512	† Signetics	(Continued)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift $\mu V/^\circ C$	Bandwidth MHz	Slew Rate V/ $\mu s$	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Dual Units</b>										<b>(Cont'd)</b>		
3.0	15	5	—	0.25	0.16	100K	—	0	Low Power	LH2250 LH2250C	† National	(Cont'd)
				25 *	0.16 *	100K	70	0	Programmable Dual LM4250	LH24250	† National	
	20	7.5	5 *	0.27 *	0.16 *	50K	70	0	Low Power Adjustable Current	ICL8022M	† Intersil	
	80	15	15	1 *	—	50K	70	0	Low Power, Single Supply	LM258A CA258A	National RCA	5
		30	15	0.44	0.3	50K	80	0	Dual 741	$\mu A747AM$ $\mu A747EC$ LM747A LM747E	† Fairchild Fairchild † National National	10
	100	30	8 *	8 *	2	25K	80	0	High Slew Rate	NA-2650	† Harris (3414)	
			20	1 *	—	25K	65	0	Low Power, Single Supply	LM358A CA358A	National RCA	
	500	200	—	4	3 *	—	—	—	High Performance, Low Noise	RC2041	Raytheon (3592)	
	750	400	3 *	—	2 *	20K	70	3	Audio Preamp	$\mu A749C$	Fairchild	15
	1000	200	—	8	6 *	—	—	—	High Performance, Low Noise	RC2043	Raytheon (3592)	
	5000	700	5 *	9 *	125 *	50K	80	1	Transconductance Amp	CA3280	RCA (3597)	
3.5	300	50	5 *	1.5	1 *	—	86	0	High Performance	TEB1033	Thomson-CSF	
4.0	80	20	4 *	1 *	10	50K	70	0	High Slew Rate	SE5535	† Signetics	20
	150	25	—	—	—	50K	70	0	2 Op Amp/Comparators	LM192	† National	
	200	50	20	5.4 *	15 *	200K	96 *	0	Precision Bipolar-JFET	OP-215C OP-215G	† PMI PMI	
	350	100	3 *	—	0.5 *	50K	80	0	Dual 222	TBC0747	† Siemens	
	500	600	—	2 *	50 *	20K *	80	0	Transconductance Amplifier	LM13700	† National	
	700	100	25	—	—	18K	70	0	20 mA Output	TAA2762	† Siemens	25
				—	—	18K	70	0	General Purpose, OC Output	TEC1761M	† Thomson-CSF	
	800	150	—	10 *	9	25K	70	1	Low Noise	NE5532 NE5532A	TI TI	
	1500	300	—	10 *	13 *	25K	70	1	Dual 5534	XR5532 XR5532A XR5533 XR5533A NE5532 NE5532A NE5533	Exar (3378) Exar (3378) Exar (3379) Exar (3379) Signetics (3644) Signetics (3644,3644) † Signetics (3646)	30
			5 *	10 *	13 *	25K	70	3	Can be compensated with 1 capacitor, but reduces BW and slew rate.	NE5533A	Signetics (3646)	35
5.0	0.03	0.02	6 *	4 *	10 *	50K *	70	0	Dual 3160	CA3260AE CA3260AT	RCA (3595) † RCA (3595)	
	0.04	0.02	15 *	45 *	9 *	20K	70	0	Dual 3140A, MOS FET	CA3240A	RCA (3595)	
	0.050	0.030	15 *	1	1.6	80K	70	0	Low Power, Dual 1458	ICL7621BC ICL7621BM	Intersil (3481) † Intersil (3481)	40
									Low Power, Dual 747	ICL7622BC ICL7622BM	Intersil (3481) † Intersil (3481)	
	0.10	0.05	10 *	1 *	1 *	25K	70	0	Low Power JFET	LF442	National	45
		0.10	10 *	3 *	13 *	50K	80	0	Bipolar JFET	$\mu A772B$ $\mu A772BM$	Fairchild † Fairchild	
	0.2	0.1	10 *	4 *	13 *	25K	80	0	Bipolar-JFET	LF353B	Harris	
			30	4 *	13 *	50K	80	0	Dual 0351B	TDB0353B	Thomson-CSF	
	0.20	0.1	10 *	4 *	13 *	25K	70	0	Bipolar-JFET	MC35002B	† Motorola	50
		0.10	10 *	3	10	50K	80	0	Dual 151	LF153	† National	
				4 *	13 *	50K	80	0	Dual 351B	LF353B	National	
	7.5	3	3 *	1 *	15 *	200K	70	0	Multi-Purpose Programmable	$\mu A776M$	† Fairchild	
	10	5	10 *	—	0.1 *	25K	74	0	Programmable	HA-2735	Harris	
		10	—	1	1	100K	100	0	Low Noise	NE5512 NE5517 NE5517A	Signetics Signetics (3642) Signetics (3642)	55
	100	25	10	0.8	0.25	50K	70	0	Dual Matched	OP-04B OP-04D	† PMI (3566) PMI (3566)	
			20	0.8	0.25	25K	70	0	Dual Matched, High Performance	OP-14D	PMI (3566)	
						50K	70	0	Dual Matched	MPOP-04B	† MicroPwr (3529)	60

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Dual Units</b>										<b>(Cont'd)</b>			
5.0	100	25	20	0.8	0.25	50K	70	0	Dual Matched	MPOP-04B MPOP-14B MPOP-14B	MicroPwr (3529) † MicroPwr (3529) MicroPwr (3529)		
									MPOP-04 With Internally Connected Vcc Terminal	MPOP-03B MPOP-03D	MicroPwr (3529) MicroPwr (3529)	5	
		40	—	0.8*	0.5*	4K	60	0	Low Power	TL022M	† TI		
		50	20	0.8	0.25	25K	70	0	Dual Matched, High Performance	OP-14B	PMI (3586)		
150	30	7*	1*	—	—	50K	70	0	Half LM124	LM158 LM158 LM258 CA158 CA258 LM158 LM258 SE532 LM158 TDC0158	† Motorola † National National † RCA RCA † Signetics Signetics † Signetics † TI † Thomson-CSF	10 15	
200	60	8*	8*	2	20K	74	0	High Slew Rate	HA-2655	Harris			
	80	—	0.8*	0.5*	1K	60	0	Low Power	TL022C	TI			
250	50	—	—	—	25K	65	0	2 Op Amp/Comparators	LM292 LM392	National National		20	
500	50	10*	1*	0.8*	50K	70	0	Single Supply	MC3558	† Motorola			
	200	—	—	10	50K	70	0	High Slew Rate 1558	MC1558S	† Motorola			
			1*	0.5*	50K	70	0	Dual 741	SFC2747M	† Thomson-CSF			
				0.8*	50K	70	0	Dual 741	SFC2458M	† Thomson-CSF		25	
				2	1.5*	50K	70	0	Dual Wideband 741	RM4558	† Raytheon		
				2.5	1	20K	70	0	General Purpose	SE4558	† Signetics		
					1.5	50K	70	0	Dual Wideband 741	TDC4558	† Thomson-CSF		
					1.5*	50K	70	0	Dual Wideband 741	MC4558 MC4558A	† Motorola Motorola		30
									Low Noise	MC4558N	† Motorola		
				3	1.5*	50K	70	0	3 MHz Min. Bandwidth	RM4559	† Raytheon		
			1.5*	1*	0.3*	25K	70	3	Matched Dual 709	RM1537	† Raytheon		
			2-15*	1*	0.5*	50K	70	0	Dual 741	μA1558M μA747M MC1558 MC1747 LM1558 LM747 PM-1558 PM-747 RM1558 RM747 CA1558 CA747 MC1558 μA747 SG1558 SG747 MC1558 μA747M TA75458 TA75747	† Fairchild † Fairchild † Motorola † Motorola † National † National † PMI † PMI † Raytheon † Raytheon † RCA † RCA † Signetics † Signetics † SiliconG † SiliconG † TI † TI Toshiba Toshiba	35 40 45 50	
									Low Noise	MC1558N	† Motorola		
				3*	0.25	25K	70	3	Dual MC1709	MC1537	† Motorola		55
1000	200	—	0.12	7*	30K	80	0	Low Noise, High Speed	LM833	National			
5000	600	7*	2*	50*	—	80	0	Transconductance Amplifier, G = 13000 μmho.	LM13600	Signetics (3636)			
7000	8000	6000	—	2*	50	80	0	Transconductance Amplifier	LM13600	National			
6.0	0.04	0.02	10*	5*	13*	50K	80	0	MOS/FET Input; Bipolar/MOS Output	CA082A CA082T CA083A	RCA (3596) † RCA (3596) RCA (3596)	60	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Dual Units</b>										<b>(Cont'd)</b>			
6.0	0.2	0.05 0.1	10 * 10 *	3 * 1 *	13 * 3.5 *	50K 4K	80 80	0 0	Low Noise Bipolar-JFET Low Power Bipolar-JFET	TL072M TL062AC TL062M	† TI TI † TI	(Cont'd)	
				3 *	13 *	50K	80	0	Bipolar JFET	<b>XR082</b> <b>XR082M</b> <b>XR083</b> <b>XR083M</b> TL082A TL082AC TL082M TL083AC TL083M	Exar † Exar Exar † Exar Motorola TI † TI TI † TI	(3376) (3376) (3376) (3376)	5
				4 *	13 *	25K	80	0	JFET Input	TL082M	† Motorola		10
0.20	0.05		10 *	4 *	10 13 *	35K 50K	70 80	0 0	Low Noise Bipolar JFET Low Noise Bipolar-JFET	TL072M TL072AC TL072AC	† Motorola Motorola TI		15
10	6		3 *	1 *	15 *	50K	70	0	Multi-Purpose Programmable	μA776C	Fairchild		20
30	10		—	0.25 * 5 *	0.16 * 0.27 *	75K 50K	70 70	0 0	Programmable Dual LM4250 Low Power Adjustable Current	LH24250C ICL8022C	National Intersil		25
75	10		—	0.4	1.5	100K	80	—	Ultra-Low Power	HA-5142A	Harris		30
150	40		6 *	1 *	10	50K	70	0	High Slew Rate	NE5535	Signetics		35
250	50		7 *	1 *	0.6 *	25K	65	0	Half LM244/324	LM258 LM358 NE532	Motorola Motorola Signetics		40
	75		10 *	1 *	0.6 *	20K	70	0	Single Supply, I/O Operates to Ground	μA798C	Fairchild		45
500	200		—	—	0.5 *	5K	70	0	Dual 221	TBB0747 TBB1458	Siemens Siemens		50
				1 *		20K	70	0	Wideband 741	μPC4557	NEC		55
				2 *		20K	70	0	Wideband 741	μPC4559	NEC		60
				5 *		20K	70	0	Wideband 741	μPC4556	NEC		65
				1 *	0.5 *	20-50K	70	0	Dual 741C	IC900 XR1458 μA1458C μA747C MC1458 MC1747C LM1458 LM747C μPC251 AN6550 AN6551 AN6552 RC747 CA747C μA747C SG747C μA747C	Cherry Exar Fairchild Fairchild Motorola Motorola National National NEC Panasonic (3556) Panasonic (3556) Panasonic (3556) Raytheon RCA Signetics SiliconG TI		70
						25K	70	0	Dual 741	SFC2747C	Thomson-CSF		75
				0.8 *		20K	70	0	Dual 741	SFC2458C	Thomson-CSF		80
				2	0.5 *	20K	70	0	Dual Wideband 741C Low Noise	MC4558C MC4558NC	Motorola Motorola		85
				1	20K 50K	70 70	0 0	0	Dual Wideband 741 General Purpose	TDB4558 NE4558	Thomson-CSF Signetics		90
				3	1.5	20K	70	0	3 MHz Min. Bandwidth	RC4559	Raytheon (3592)		95
				3 *	1	20K	70	0	Low Noise Wideband 741	RC4739 XR4558 μPC4558 RC4558 RC4558	Raytheon (3592) Exar NEC Raytheon (3592) TI		100
									Low Noise	XR4739	Exar (3378)		105
				10 *	2.8 *	20K	70	0	Wideband 741	μPC4560	NEC		110
				10 *	1 *	0.5 *	25K	70	Automotive Temperature Range (-40°C to 85°C)	SA747C	Signetics	(Continued)	115

† Military Temperature Range (-55° to 125°C)

\* Typical Value  
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Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Dual Units</b>										<b>(Cont'd)</b>		
6.0	500	200	10 *	1 *	0.8 *	20K	70	0	Low Noise	MC1458N	Motorola	(Cont'd)
					10	20K	70	0	High Slew Rate 1458	MC1458S	Motorola	
			12 *	1 *	0.8 *	20K	70	0	Automotive Temperature Range (-40°C to 85°C)	SA1458	Signetics	
	1000	300	6 *	—	9 *	10K	65	0	General Purpose, OC Output	TEC1761C	Thomson-CSF	
	2000	1000	—	—	1 *	6.5K	70	2	Low Noise	μA739	Fairchild	5
7.0	100	10	—	0.4	1	100K	77	—	Ultra-Low Power Single Supply	HA-5142	Harris	
	200	50	—	—	—	25K	65	0	Single Supply	TL321C	TI	
	250	50	7 *	—	—	100K *	50	0	Single Supply Half LM224	TDF2904	Thomson-CSF	
				1 *	—	25K	65	0	Automotive Temperature Range (-40°C to 85°C)	μPC1251	NEC	10
					0.3 *	25K	65	0	Automotive Temperature Range (-40°C to 85°C)	μPC358	NEC	
										TDB0158	Thomson-CSF	
7.5	7	1	30	1 *	0.3 *	25K	80	1	Dual 308	LH2308	National	
										PM-2308	PMI	
	250	50	30	1 *	0.5 *	25K	70	1	Dual High Performance	LH2301A	Intersil	15
										LH2301A	National	
	1500	500	1.5 *	1 *	0.25 *	15K	65	3	Matched Dual MC1709C	MC1437	Motorola	
					0.3 *	15K	65	3	Matched Dual 709	RC1437	Raytheon	
8.0	500	75	10 *	1 *	0.6 *	20K	70	0	Low Power	TL322M	TI	
9.0	0.40	0.20	10 *	1 *	0.6 *	200K	70	0	JFET Input	TL092M	† TI	20
10.0	0.10	0.03	10	3	10	100K *	100 *	1	JFET Input	MA336	AnalogSys	
	0.2	0.1	10 *	4 *	13 *	25K	70	0	Dual 0351	TDB0353	Thomson-CSF	
	0.20	0.05	10 *	3 *	13 *	25K	70	0	Low Noise Bipolar-FET	MCE72	MCE	
				4 *	13 *	25K	70	0	Low Noise Bipolar-FET	TL072C	Motorola	
						50K	80	0	Low Noise Bipolar-JFET	TL072C	TI	25
	0.10	10 *	3 *	13 *	50K	70	0	Bipolar JFET	μA772	Fairchild		
			4 *	13 *	25K	70	0	Dual 351, Wide Band	LF353	Motorola		
									LF353	National		
	500	75	10 *	1 *	0.6 *	20K	70	0	Low Power	TL322C	TI	30
	700	300	15 *	1.1 *	0.5 *	20K	60	0	General Purpose	MC1458C	Motorola	
	1500	600	—	—	—	10K	70	0	Audio Preamp	μA749D	Fairchild	
15.0	0.050	0.030	8 *	4 *	10 *	50K *	70	0	Dual 3160	CA3260E	RCA (3595)	
										CA3260T	† RCA (3595)	
			10 *	5 *	13 *	50K	70	0	MOS/FET Input, Bipolar/MOS Output	CA082	RCA (3596)	35
										CA083	RCA (3596)	
						200K	76 *	0	MOS/FET Input, Bipolar/MOS Output	CA082C	RCA (3596)	
										CA083C	RCA (3596)	
			25 *	1 *	1.6	80K *	70	0	Low Power, Dual 1458	ICL7621DC	Intersil (3481)	
									Low Power, Dual 747	ICL7622DC	Intersil (3481)	40
	0.20	0.10	10 *	3 *	13 *	50K	70	0	Bipolar JFET	μA772L	Fairchild	
	0.4	0.2	10 *	1 *	3.5 *	3K	70	0	Low Power Bipolar-JFET	TL062C	TI	
				3 *	11 *	25K	70	0	Bipolar-JFET	μPC4082	NEC	
					13 *	25K	70	0	Bipolar-JFET	XR082C	Exar (3376)	
										XR083C	Exar (3376)	45
										MCE82	MCE	
										MCE83	MCE	
										TL082	Motorola	
										TL082C	TI	
										TL083C	TI	50
	50	25	12 *	—	18 *	3K	65	0	Darlington Input	TBB2331	Siemens	
	200	100	30	—	2	10K	40	0	Power Op Amp, Comp. for G>30	MA208	AnalogSys	
20.0	0.02	0.0005 *	75	1 *	6 *	50K	70	0	FET Input Dual	ICL8043M	† Intersil	
50.0	0.05	0.0005 *	75	1 *	6 *	20K	70	0	FET Input Dual	ICL8043C	Intersil	
<b>Triple Units</b>												
5.0	0.050	0.030	10 *	1	1.6	80K *	70	0	Low Power	ICL7631BC	Intersil (3481)	55
										ICL7631BM	† Intersil (3481)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line	
<b>Triple Units</b>										<b>(Cont'd)</b>			
5.0	0.050	0.030	10 *	1	1.6	80K *	70	1	Low Power	ICL7632BC	Intersil (3481)	(Cont'd)	
										ICL7632BM	† Intersil (3481)		
	70	14	—	0.02 *	0.1 *	—	70	0	Micropower, Transconductance Amplifier	CA3060	† RCA		
10.0	0.050	0.030	20 *	1	1.6	80K	70	0	Low Power	ICL7631CC	Intersil (3481)	5	
										ICL7631CM	† Intersil (3481)		
								1	Low Power	ICL7632CC	Intersil (3481)		
										ICL7632CM	† Intersil (3481)		
	2000	200 *	—	5	0.46	4K *	90 *	2	Supply +6 V	TCA220	Signetics		
20.0	0.050	0.030	30 *	1	1.6	80K	70	0	Low Power	ICL7631EC	Intersil (3481)	10	
								1	Low Power	ICL7632EC	Intersil (3481)		
<b>Quad Units</b>													
										HA4741-8	Raytheon		
0.5	0.075	0.025	10 *	4 *	13 *	50K	80	0	Precision BIFET	MC35004A	† Motorola	15	
	300	20	10	1.5	0.7	100K	100	0	Symmetrical, Matched	MP5509A	† MicroPwr (3529)		
										MP5509E	MicroPwr (3529)		
										MP5511A	† MicroPwr (3529)		
										MP5511E	MicroPwr (3529)		
										MPOP-11	† MicroPwr		
										OP-09A	† PMI (3566)		
										OP-09E	PMI (3566)		
										OP-11A	† PMI (3566)	20	
										OP-11E	PMI (3566)		
2.0	0.2	0.1	8.3 *	0.24	15 *	50K	80	—	JFET Input	HA-5084B-5	Harris (3438)	25	
			10 *	4 *	13 *	25K	80	0	Bipolar-JFET	LF347A	Harris		
		2	10 *	4 *	13 *	50K	80	0	High Speed, JFET Input	TDB0347A	Thomson-CSF		
	50	10	20	1 *	—	50K	70	0	Low Power	LM124A	† National		
	500	50	10 *	3 *	8	50K	80	0	High Performance Bipolar	MC33074A	Motorola (3530)		
										MC34074A	Motorola		
										MC35074A	† Motorola		
2.5	50	5	10	750 *	0.25 *	400K *	100K	0	Quad Micropower OP-21	OP-421B	† PMI (3566)	30	
										OP-421F	PMI (3566)		
	200	75	2 *	0.26	12	100K	86	0	Wideband, High Performance	HA-4620-2	† Harris (3428)	35	
										HA-4620-5	Harris (3428)		
										HA-4600-2	† Harris (3426)		
										HA-4600-5	Harris (3426)		
										HS4602RH	† Harris (4721)		
						4 *	100K	86	0	Wideband	HA-4602-2	† Harris	
			3 *	8 *	1	100K	86	0	Low Noise, High Performance	HA-5104-2	† Harris (3459)	40	
										HA-5104-5	Harris (3459)		
						60 *	12	100K	86	1	Wideband	HA-5114-2	† Harris (3459)
										HA-5114-5	Harris (3459)		
	500	50	15	1.5	0.7	100K	100	0	Symmetrical, Matched	OP-09B	† PMI (3566)	45	
										OP-09F	PMI (3566)		
										OP-11B	† PMI (3566)		
										OP-11F	PMI (3566)		
3.0	0.03	0.02	10 *	5 *	13 *	50K	80	0	MOS/FET Input, Bipolar/MOS Output	CA084B	RCA (3596)	50	
	0.2	0.1	10 *	0.63	2	20K	80	—	Low Power, JFET Input	HA-5064B-5	Harris (3434)		
						1 *	3.5 *	4K	80	0	Low Power Bipolar-JFET	TL064BC	TI
						3 *	3	50K	80	0	JFET	TL084BC	TI
	0.20	0.05	10 *	4 *	13 *	25K	70	0	Low Noise Bipolar-JFET	TL074BC	TI		
	20	10	—	1 *	0.8 *	30K	100 *	0	Programmable	HA-2740-2	† Harris (3422)		
	80	15	20	1 *	—	50K	70	0	Low Power	LM224A	National		
	100	30	30	—	—	100K	65	0	High Gain, Single Supply	TDB0124A	Thomson-CSF		
				1 *	—	25K	65	0	Low Power	LM324A	National		
										LM324A	TI		
	200	30	5 *	3	1.2	25K	80	0	Noise 2μV Max	RM4156	† Raytheon	55	
					1.6 *	50K	80	0	Quad 741	XB4741M	† Exar (3378)		
					3.5 *	1.6 *	50K	80	0	Quad 741	HA-4741-2		† Harris (3430)
										HA4741-2	† Raytheon	(Continued)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Quad Units</b>										<b>(Cont'd)</b>		
3.0	200	30	5 *	19 *	8 *	25K	80	0	Uncompensated 4156	RC4157 RM4157	Raytheon † Raytheon	(Cont'd)
	325	125	2 *	70 *	12	100K	86	0	Wideband, Comp. for g > 10	HA-4622-2	† Harris	
3.5	300	100	2 *	8 *	4 *	75K	80	0	Wideband	HA-4605-5	Harris	
4.0	125	12	15	750 *	0.25 *	200K	96 *	0	Quad, Micropower OP-21	OP-421C OP-421G	† PMI PMI	(3566) (3566)
	400	50	5 *	1 *	1.2 *	50K	70	0	LM124 with Improved Output	RM3503A	† Raytheon	
		120	2 *	70 *	11	75K	80	0	Wideband, Comp. for g > 10	HA-4625-5	Harris	
4.5	500	75	10 *	3 *	10	25K	70	0	High Performance Bipolar	MC33074 MC34074 MC35074	Motorola † Motorola † Motorola	(3530)
5.0	0.050	0.030	15 *	1	1.6	80K	70	0	Low Power	ICL7641BC ICL7641BM ICL7642BC ICL7642BM	Intersil † Intersil Intersil † Intersil	(3481) (3481) (3481) (3481)
	0.050	10	1	1	50K	80	0	Low Power BIFET	LF444A	National		
0.10	0.05	10 *	4 *	13 *	50K	80	0	Precision BIFET	MC35004B	† Motorola		
	0.10	10 *	3 *	13 *	50K	80	0	Bipolar JFET	μA774B μA774BM	Fairchild † Fairchild		
0.2	0.1	10 *	4 *	13 *	25K	80	0	Bipolar-JFET	LF347B	Harris		
	4	10 *	4 *	13 *	50K	80	0	High Speed, JFET Input	TDB0347B	Thomson-CSF		
0.20	0.10	8.3 *	0.24	15 *	50K	80	—	JFET Input	HA-5084-2 HA-5084A-5	† Harris Harris	(3438) (3438)	
		10 *	4 *	13 *	50K	80	0	Precision BIFET Wideband, Quad 351B	MC34004B LF347B LF347B	Motorola Motorola National		
30	10	—	1 *	0.8 *	25K	74	0	Programmable	HA-2740-5	Harris	(3422)	
100	20	—	0.8	0.4 *	1M	70	0	High Gain, Programmable	TDC0146	† Thomson-CSF		
	25	—	1 *	0.5 *	50K	70	0	High Gain, Programmable	TDC0148 TDC0149	† Thomson-CSF † Thomson-CSF		
								Quad 741 with Standard npn Input Stage	LM148 μA148 LM148 LM148 LM148	† AMD † Fairchild † Motorola † Raytheon † TI		
			4 *	2.0 *	50K	70	0	Wideband Quad 741 for Gains > 5	LM149	† National		
	40	—	0.8 *	0.5 *	4K	60	0	Low Power	TL044M	† TI		
150	30	7 *	—	—	100K	70	0	High Gain, Single Supply	TDC0124	† Thomson-CSF		
250	80	—	0.8 *	0.5 *	1K	60	0	Low Power	TL044C	TI		
300	50	5 *	0.02	1.3	50K	80	0	General Purpose	HA-4156-5	Harris	(3424)	
			3	1.2	25K	80	0	Noise 2 μV Max	RC4156	Raytheon		
			3 *	1.6 *	25K	80	0	Quad 741	μPC4741	NEC		
			3.5 *	1.6 *	25K	80	0	Quad 741	XR4741C HA-4741-5 HA4741-5	Exar Harris Raytheon	(3378) (3430)	
500	50	—	2 *	1.5 *	5K	70	0	Programmable Quad 741	MC4202	Motorola		
				1.6 *	20K	70	0	Low Noise, Low Power	XR4212M	† Exar	(3378)	
			10	1	200K	90	0	Low Power	MC3503	Signetics		
			10 *	1 *	0.6 *	—	70	0	Low Power, Single Supply	TDC3403	† Thomson-CSF	
	200	—	3 *	1.5 *	50K	70	0	Quad 741, High Gain	XR4136M μA4136M MP4136 RM4136	† Exar † Fairchild MicroPwr † Raytheon	(3378) (3529)	
			3.5	1.5	96	110	0	Programmable	TAB1042 TAB1043	Plessey Plessey		
			4 *	1.5	0.7	50K	70	0	Quad Matched 741	OP-09C	PMI	(3566)
			15 *	1	0.5 *	50K	70	0	Quad 741	MC4741	† Motorola	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line																						
<b>Quad Units</b>										<b>(Cont'd)</b>																								
6.0	0.04	0.02	10 *	5 *	13 *	50K	80	0	MOS/FET Input, Bipolar/MOS Output	CA084A	RCA (3596)	5																						
			0.05	8	20	50K	80	0	FET, Wideband	MC3571	† Motorola																							
			10 *	3 *	13 *	50K	80	0	Low Noise Bipolar-JFET	MCE74	MCE																							
	0.1	10 *	0.63	2	20K	80	—	0	—	Low Power, JFET Input	NA-5064-2		† Harris (3434)																					
											NA-5064A-5		Harris (3434)																					
											TL064AC		TI																					
											1 *		3.5 *	4K	80	0	—	0	—	—	TL064AC	TI												
																					3 *	13 *	50K	80	0	Bipolar JFET	XR084	Exar (3377)						
											XR084M		† Exar	(3377)	MCE84	MCE	TL084A	Motorola	TL084AC	TI	TL084M	† TI												
																							Indiv. Program	XR096	Exar (3377)									
XR096M	† Exar (3377)																																	
Programmable, Bipolar - FJET	XR094	Exar (3377)																																
	XR094M	† Exar (3377)																																
	XR095	Exar (3377)																																
XR095M	† Exar (3377)																																	
0.20	0.05	10 *	4 *	13 *	25K	70	0	—	Low Noise Bipolar-JFET	TL074AC	TI																							
										50K	80	0	Low Noise Bipolar JFET	TL074BC	Motorola																			
										—	—	—	Low Noise Bipolar-JFET	TL074AC	Motorola																			
75	10	—	0.4	1.5	100K	80	—	Ultra-Low Power	HA-5144A	Harris																								
200	0.1	10 *	3 *	13 *	50K	80	0	—	High Speed, JFET Input	TDC0084	Thomson-CSF																							
										50	—	1 *	0.5 *	25K	70	0	High Gain, Programmable	TDB0148	Thomson-CSF															
Quad 741 with Standard npn Input Stage	—	—	—	—	—	—	—	—	—	TDB0149	Thomson-CSF																							
										LM248	AMD																							
												LM348	AMD																					
														μA248	Fairchild																			
																μA348	Fairchild																	
																		LM248	National															
																				LM348	National													
																						LM248	Raytheon											
																								LM348	Raytheon									
																										LM248	TI							
LM348	TI																																	
		4 *	2 *	25K	70	0	—	—	—	Wideband Quad for Gains > 5	LM249																	National						
											LM349	National																						
											LM249	Raytheon																						
											LM349	Raytheon																						
											250	20	25	750 *	0.25 *	200K *	90 *	0	—									Quad OP-21	OP-421H	PMI (3566)				
																				100	—								0.5	0.4 *	1M	70	0	High Gain, Programmable
											500	50	—	1 *	1.2 *	25K	70	0	—	LM324 with Improved Output	RC3403A	† Raytheon												
																					2	1.6	5K	70	0			—	—	Quad 741	XR4212C	Exar (3378)		
																										100	—				1	0.4 *	20K	70
200	—										1 *	0.5 *	20K	70	0	—	Quad 741	MC4741C	Motorola															
		3 *	1 *	20K	60-70	0	—	—	—	Quad 741								XR4136C	Exar (3378)															
μA4136C	Fairchild																																	
		MP4136C	MicroPwr (3529)																															
RC4136	Raytheon																																	
		SG4136	† SiliconG																															
RC4136C	SiliconG																																	
		RC4136	† TI																															
2.4 *	20K			70	0	—	—	—	—	High Performance	TL136C	TI																						
		7	250								50	7 *	—	—	100K *	50	0	Single Supply	TDF2902	Thomson-CSF														
7.0	100	10	—	0.4	1	100K	77	—	—	Ultra-Low Power	HA-5144	Harris																						
											150	30	—	—	—	—	—	—	Single Supply	TA75902	Toshiba													
	250	50	—	—	—	—	—	—	—	25K										65	0	—	—	—	324 Type	TBB0324	Siemens							
											7 *	—	—	—	—	—	—	—	100K							65	0	—	—	—	—	High Gain, Single Supply	TDB0124	Thomson-CSF
																																	1 *	0.3 *
500	50	—	—	—	—	—	—	—	100K	50	0	—	—	—	—	—	Single Supply	μPC2902	NEC															
																		100K *	85 *	0	—	—	—	—	—	—	μPC451	NEC						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Operational Amplifiers-Characteristics (Cont'd)

Offset Voltage mV (25°C)	Bias Current nA (25°C)	Offset Current nA (25°C)	Voltage Drift μV/°C	Bandwidth MHz	Slew Rate V/μs	Voltage Gain V/V	CMRR dB	Comp.	Comments	Device	Source	Line
<b>Quad Units</b>										<b>(Cont'd)</b>		
8.0	500	75	10 10 *	1 1 *	0.6 0.6 *	200K 20K	90 70	0 0	Low Power Single Supply: I/O Operates to Ground	MC3303 MC3303 TDF3403	Signetics TI Thomson-CSF	
9.0	0.2	0.1	10 *	1 * 4 *	3.5 * 13 *	4K 25K	80 80	0 0	Low Power Bipolar-JFET JFET Input	TL064M TL084M	† TI † Motorola	5
	0.20	0.05	10 *	4 *	10	35K	70	0	Low Noise Bipolar JFET	TL074M	† Motorola	
	0.40	0.20	10 *	1 *	0.6 *	200K	70	0	JFET Input	TL094M	† TI	
10.0	0.050	0.030	20 *	1 *	1.6 *	80K	70	0	Low Power	ICL7641CC ICL7641CM ICL7642CC ICL7642CM	Intersil (3481) † Intersil (3481) Intersil (3481) † Intersil (3481)	10
	0.10	0.05	10	1	1	25K	70	0	Low Power BIFET	LF444	National	
	0.2	0.1	10 *	4 *	13 *	25K	70	0	Bipolar-JFET	LF147 LF247 MC34004 MC35004	Harris Harris Motorola † Motorola	15
		4	10 *	4 *	13 *	25K	70	0	High Speed, JFET Input	TDB0347	Thomson-CSF	
	0.20	0.05	10 *	4 *	13 *	25K	70	0	Low Noise Bipolar-JFET	TL074C TL074C TL075C	Motorola TI TI	20
		0.1	10 *	4 *	13 *	25K	70	0	Bipolar-JFET, Quad 351	LF347 LF347	Motorola National	
		0.10	10 *	3 *	13 *	50K	70	0	Bipolar JFET	μA774	Fairchild	
	0.20 *	0.05 *	10 *	1 *	0.6 *	20K	70	0	JFET Input	TL094M	† TI	
	500	50	10 10 *	1 1 *	0.6 0.6 *	200K — 20K	90 70 70	0 0 0	Low Power Low Power, Single Supply Single Supply: I/O Operates to Ground	MC3403 TDB3403 μA3403 MC3403	Signetics Thomson-CSF Fairchild TI	25
15.0	0.05	0.03	10 *	5 *	13 *	25K	70	0	MOS/FET Input, Bipolar/MOS Output	CA084	RCA (3596)	
	0.20	0.10	10 *	3 *	13 *	50K	70	0	Bipolar-JFET	μA774L	Fairchild	30
	0.40	0.2	8.3 * 10 *	0.24 1 * 3 *	15 * 3.5 * 11 * 13 *	25K 3K 25K 25K	70 70 70 70	— 0 0 0	JFET Input Low Power Bipolar-JFET Bipolar-FET Bipolar-FET	HA-5084-5 TL064C μPC4084 XR084C TL084 TL084C TL085C	Harris (3438) TI NEC Exar (3377) Motorola TI TI	35
									Indiv. Program.	XR096C	Exar (3377)	
									Programmable Bipolar-JFET	XR094C XR095C	Exar (3377) Exar (3377)	40
			20 *	0.63	2	10K	70	—	Low Power, JFET Input	HA-5064-5	Harris (3434)	
	50	25	12 *	—	—	3K	65	0	Darlington Input	TBB4331 TBE4335	Siemens Siemens	
	400	0.2	10 *	3 *	13 *	25K	70	0	High Speed, JFET Input	TDB0084	Thomson-CSF	
20.0	0.050	0.030	30 *	1	1.6	80K	70	0	Low Power	ICL7641EC ICL7642EC	Intersil (3481) Intersil (3481)	45
30.0	0.050	0.100	15 *	1 *	0.8 *	1K *	70 *	0	Low Power	MC14573	Motorola	

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LINEAR-Phase Locked Loops/Synthesizers

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Phase Locked Loops/Synthesizers</b>				<b>Phase Locked Loop System (multiplier, VCO and Op Amp)</b>				<b>Synthesizer, PLL</b>			
CB device sets and support circuits are also found in Consumer Circuits				XRS200 Exar				HCTR0347 Hughes 95			
Count Extender, +4, +10/11 becomes +40/41				Phase-Frequency Detector				MC145104 Motorola			
SP8790A † Plessey				MC12040 Motorola				MC145106 Motorola			
SP8790B Plessey				MC12540 † Motorola 55				MC145107 Motorola			
Count Extender, +8 (extends division ratio of a two modulus prescaler while retaining the difference, ie R10/11 becomes +80/81)				MC4044 Motorola				MC145109 Motorola			
SP8794A † Plessey				MC4344 † Motorola				MC145112 Motorola 100			
SP8794B Plessey				PLL Peripheral (VCO, mixer, amp for CB)				MC145143 Motorola			
Counter Logic Control				AN103 Panasonic				MC145144 Motorola			
SP8790 Plessey				TA7310 Toshiba				MC145145 Motorola			
Counter Logic Control (use with MC12012 for high frequency programming)				Prescaler SDA4041 Siemens 60				MC145146 Motorola			
MC12014 Motorola				Prescaler, Divide by 20				MC145151 Motorola 105			
Crystal Oscillator (2 to 20 MHz)				MC3396 Motorola				MC145152 Motorola			
MC12061 Motorola				Prescaler, Divide by 64				MC145155 Motorola			
MC12561 † Motorola				SDA2001 Siemens				MC145156 Motorola			
Crystal Oscillator (250 kHz to 60 MHz)				Prescaler, for AM/FM				MC145157 Motorola			
CK1100A Solarise				TD6102 Toshiba				MC145158 Motorola 110			
CK1114A Solarise				Prescaler, for FM Radio				MC145159 Motorola			
CK1144A Solarise				DS8626 National				MC6195 Motorola			
CK1145A Solarise				DS8629 National 65				DS8906 National			
FSK Modulator/Demodulator				SP8629 Plessey				DS8907 National 115			
XR210 Exar (3365)				Prescaler for PLL Tuner, Divide by 8				DS8908 National			
XR210M † Exar (3365)				LB3500 Sanyo				MM55108 National			
XR2211 † Exar				Prescaler for VHF/UHF, Divide by 64				MM55110 National			
XR2211C Exar				MC12071 Motorola				MM55121 National			
XR2211M † Exar				DS8621 National				MM55123 National			
XR2212M † Exar (3366)				CA3179 RCA 70				MM55124 National			
XR2211C Raytheon				Prescaler for VHF/UHF, Divide by 4				MM55126 National			
XR2211M † Raytheon				CA3199 RCA (3603)				MN6040 Panasonic			
Phase Comparator and Programmable Counter				Prescaler for VHF/UHF, Dual Modulus				MN6040A Panasonic			
MC145688A † Motorola				DS8622 National				MN6040Z Panasonic			
MC145688C Motorola				Prescaler, Low Power (+64), 1.1 GHz				MG142 Panasonic			
Phase Locked Loop				MC12073 Motorola				LC7210 Sanyo			
XR215 Exar (3365)				Prescaler, Low Power (+64), 225 MHz				LC7220 Sanyo			
XR215M † Exar (3365)				MC12023 Motorola				LC7225 Sanyo			
SL652 Plessey				Prescaler, Low Power (+256), 1.1 GHz				NEF4750 Signetics (865,899) 130			
SDA2002 Siemens				MC12074 Motorola				SAA1057 Signetics			
SDA2112-2 Siemens				MC12075 Motorola 75				TC9106B Toshiba			
NE584 Signetics (3649)				Prescaler, UHF (+2), 740 MHz				TC9109B Toshiba			
SE564 † Signetics (3649)				MC12090 Motorola				TC9111B Toshiba			
SN54LS297 † TI (1015)				Prescaler, 0.05 to 1.3 GHz, Divide by 256				Synthesizer (programmable divider, to 1021 channels, adder, phase comparator)			
SN74LS297 TI (1015)				SP4742 Plessey				HCTR0320 Hughes 135			
Phase Locked Loop (CMOS)				Prescaler, 0.05 to 1.3 GHz, Divide by 4056 and 8192				Synthesizer (programmable dividers, phase comparator)			
F4046BC Fairchild				SP4780 Plessey				S187A Siemens			
F4046BM † Fairchild				Prescaler, 0.08 to 1 GHz, Divide by 256				Synthesizer, 200 Channel, for AM/FM			
MC14046BA † Motorola				SP4642 Plessey 80				8X08 Signetics			
MC14046BC Motorola				SP4652 Plessey				Tone Decoder			
CD4046BC National				SP4653 Plessey				XR2211C Exar			
CD4046BM † National				Prescaler, 0.08 to 1 GHz, Divide by 64				XR2211M † Exar (3366) 140			
CD4046A † RCA (840)				SP4632 Plessey				XR2213C Exar (3366)			
CD4046AE RCA (840)				Prescaler, 1 GHz, Divide by 256				XR2213M † Exar (3366)			
CD4046B † RCA (840)				CA3211 RCA (3603)				XR567AC Exar (3367)			
CD4046BE RCA (840)				SAB1256 Signetics				XR567AM † Exar (3367)			
CD54HCT297 † RCA				Prescaler, 1 GHz, Divide by 64				XRL567C Exar (3367)			
CD74HCT297 RCA (842)				SAB1164 Signetics				XRL567M † Exar (3367) 145			
74HCT297 Signetics (907)				Prescalers—see also Digital-Special, Dividers, Digital-ECL				LM567 † National			
74HCT4046 Signetics (908)				Programmer, for CB Synthesizer				LM567C National			
883/4046B † SSS				MP7156 MicroPwr				XR2211C Raytheon			
883/4446B † SSS				MM57190 National				XR2211M † Raytheon			
SCL4046B SSS				Programmings and Controllers for Citizens Band—See Linear-Consumer Circuits				NE567 Signetics			
SCL4446B SSS				Synthesizer, Mobile Radio (2 device set)				SE567 † Signetics 150			
Phase Locked Loop for Motor Control—See Motor Speed Regulators in Linear-Consumer Circuits, Miscellaneous.				NJ8811 Plessey				VCO			
				NJ8812 Plessey				MC1648 Motorola			
				SP8901 Plessey				MC1648M † Motorola			
				SP8906 Plessey				MC1658 Motorola			
								SP1658 Plessey 155			
								SN54LS324 † TI			
								SN54LS824 † TI (1067)			
								SN54LS628 † TI (1068)			
								SN74LS824 TI (1067)			
								SN74LS828 TI (1068) 160			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page notes.

LINEAR

Master Selection Guide

LINEAR-Phase Locked Loops/Synthesizers (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	
<b>Phase Locked Loops/Synthesizers (Cont'd)</b>								
<b>VCO and Phase Comparator</b>				<b>Two Modulus Prescaler (<math>\div 5/6</math>)</b>				
	LM565	Motorola	5		MC12009	Motorola	60	
	NE565	Motorola			MC12509	† Motorola		
	LM565	National		<b>Two Modulus Prescaler (<math>\div 6/7</math>)</b>				
	LM565C	† National			SP8741	Plessey		
	NE565	Signetics		<b>Two Modulus Prescaler (<math>\div 8/9</math>)</b>				
	SE565	† Signetics			MC12011	Motorola	65	
	TA7133	Toshiba		MC12511	† Motorola			
<b>VCO Function Generator, (Includes square and triangular waveforms - but not sine.)</b>				<b>Two Modulus Prescaler (<math>\div 10/11</math>)</b>				
	XR2207	Exar	10		11C90C	Fairchild	65	
	XR2207C	Exar			11C90M	† Fairchild		
	<b>XR2209C</b>	<b>Exar (3374)</b>			MC12013	Motorola	70	
	<b>XR2209M</b>	<b>† Exar (3374)</b>			MC12513	† Motorola		
	LM566	National			SP8643	Plessey		
	LM566C	† National			SP8647	Plessey	75	
	XR2207C	Raytheon			SP8685	Plessey		
	XR2207M	† Raytheon			SP8690	Plessey		
	NE566	Signetics		15	<b>Two Modulus Prescaler (<math>\div 15/16</math>)</b>			
	SE566	† Signetics			MC3393	Motorola		
<b>VCO Phase Comparator, Multiplier</b>				<b>Two Modulus Prescaler (<math>\div 20/21</math>), 225 MHz</b>				
	SL651	Plessey		MC12019	Motorola	75		
<b>VCO Phase Comparator, Multiplier, with Auxiliary Amplifier</b>				<b>Two Modulus Prescaler (<math>\div 32/33</math>), 225 MHz</b>				
	SL650	† Plessey		MC12015	Motorola			
<b>VCO Waveform Generator (includes sinewave output among its functions)</b>				<b>Two Modulus Prescaler (<math>\div 40/41</math>)</b>				
	XR205	Exar	20		SP8793	Plessey	80	
	<b>XR2206</b>	<b>† Exar (3373)</b>			SP8793A	Plessey		
	XR2206	Exar			<b>Two Modulus Prescaler (<math>\div 40/41</math>), 225 MHz</b>			
	<b>XR2206C</b>	<b>Exar (3373)</b>			MC12016	Motorola		
	<b>XR2206M</b>	<b>† Exar (3373)</b>			<b>Two Modulus Prescaler (<math>\div 64/256</math>)</b>			
	<b>XR8038</b>	<b>Exar (3374)</b>			MC12071	Motorola	85	
	<b>XR8038A</b>	<b>Exar (3374)</b>			DS8621	National		
	<b>XR8038M</b>	<b>† Exar (3374)</b>			CA3163	RCA		
	<b>ICL8038C</b>	<b>Intersil (3518)</b>			<b>CA3179</b>	<b>RCA (3603)</b>		
	<b>ICL8038M</b>	<b>† Intersil (3518)</b>		25	<b>Two Modulus Prescaler (<math>\div 64/65</math>), 225 MHz</b>			
				MC12017	Motorola			
<b>VCO, Dual</b>				<b>Two Modulus Prescaler (<math>\div 80/81</math>)</b>				
	MC4024	Motorola	30		SP8792	Plessey	85	
	MC4324	† Motorola			SP8792A	Plessey		
	SN54LS325	TI		<b>Two Modulus Prescaler (<math>\div 128/129</math>), 225 MHz</b>				
	SN54LS326	† TI		MC12018	Motorola			
	SN54LS327	† TI		<b>Two Modulus Prescaler (<math>\div 248/256</math>)</b>				
	<b>SN54LS625</b>	<b>† TI (1067)</b>		SAB1077	Signetics			
	<b>SN54LS626</b>	<b>† TI (1067)</b>	35	<b>Two Modulus Prescaler, Low Power (<math>\div 128/129</math>), 1 GHz</b>				
	<b>SN54LS627</b>	<b>† TI (1068)</b>		MC12022	Motorola			
	<b>SN54LS629</b>	<b>† TI (1068)</b>		<b>Four Modulus Prescaler (<math>\div 256/512</math>)</b>				
	<b>SN54S124</b>	<b>† TI (962)</b>		SP8901	Plessey	90		
	SN74LS124	TI	40	SP8906	Plessey			
	SN74LS325	TI						
	SN74LS326	TI						
	SN74LS327	TI						
	SN74LS328	TI						
	<b>SN74LS625</b>	<b>TI (1067)</b>	45					
	<b>SN74LS626</b>	<b>TI (1067)</b>						
	<b>SN74LS627</b>	<b>TI (1068)</b>						
	<b>SN74LS629</b>	<b>TI (1068)</b>						
	<b>SN74S124</b>	<b>TI (962)</b>						
<b>Dual Tone Decoder</b>								
	<b>XR2567C</b>	<b>Exar (3367)</b>	50					
	<b>XR2567M</b>	<b>† Exar (3367)</b>						
<b>Two Modulus Prescaler (<math>\div 3/4</math>)</b>								
	SP8720	Plessey						
<b>Two Modulus Prescaler (<math>\div 4/8</math>)</b>								
	11C91C	Fairchild	55					
	11C91M	† Fairchild						
	MC12009	Motorola						
	MC12509	† Motorola						
	SP8740	Plessey						
	TD6102	Toshiba						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Telecommunication Circuits

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line	
<b>Telecommunication Circuits</b>												
Amplifier, Microphone	SL792	Plessey		Codec/Filter, A Law	<b>S3506</b>	<b>AMI (3348)</b>		Codec, $\mu$ 255 Law	TP5116	National	(Cont'd)	
	SL793	Plessey			MT3506	Mitel	60		TP5116A	National		
Amplifier, Telephone Handset	TEA1045	ITT			MT8961	Mitel			<b>MSM6017</b>	<b>OKI (4118)</b>	120	
Bandpass Filters	R5602A-1	Reticon			MT8963	Mitel			M5116	SGS		
	R5602A-3	Reticon	5	Codec/Filter, A Law, Asynchronous	ZNPCM4	Ferranti	65		ST101	Signetics		
	R5602A-7	Reticon		Codec/Filter, A Law or $\mu$ 255 Law metal option	<b>CD22301</b>	<b>RCA (840)</b>			TCM2910A	TI		
	R5602A-8	Reticon		Codec/Filter, A Law PCM	S3503	AMI			TCM4110	TI		
Bandpass Filters (touchtone)	S3525A	AMI			S3504	AMI			TCM4910	TI		
	S3525B	AMI		Codec/Filter, A Law, Synchronous	ZNPCM3	Ferranti	70	Companding A/D and D/A Converter	MM8100	National	125	
	<b>S3526A</b>	<b>AMI (3346)</b>	10		TP3051	National		Companding Analog to Digital Converter	DF331A	Siliconix		
	<b>S3526B</b>	<b>AMI (3346)</b>		Codec/Filter, Parallel Data Interface	TP3056	National		Companding Digital to Analog Converter	DF332A	Siliconix		
	AMS3040A	Aptek			TP3057	National		Compondor (signal expander/compressor)	XR2216	Exar		
	AMS3041A	Aptek		Codec/Filter, Pin Selectable $\mu$ Law or A Law	2913	Intel			<b>NE570</b>	<b>Signetics (3616)</b>	130	
	AMS3044	Aptek			2914	Intel			<b>NE571</b>	<b>Signetics (3616)</b>		
	AMS3045	Aptek			<b>MC14403</b>	<b>Motorola (3536)</b>	75		<b>NE572</b>	<b>Signetics (3618)</b>		
	CH1295	Cermetek			<b>MC14405</b>	<b>Motorola (3536)</b>			<b>SA571</b>	<b>Signetics (3616)</b>		
	CH1296	Cermetek			TCM2913	TI		Crosspoint Switch (double 4x4 matrix)	RC4444	Raytheon		
	MT8865	Mitel			TCM2914	TI			RM4444	Raytheon		
	AF121	National			TCM2916	TI		Crosspoint Switch (2x8 matrix)	M089	SGS	135	
	AF122	National	20		TCM2916	TI			M099	SGS		
Bandpass Filters Tunable, See also Filter, Modem Filters	R5626	Reticon			TCM2917	TI		Crosspoint Switch (4x4 matrix with control memory)	<b>MC142100</b>	<b>Motorola (3536)</b>	140	
Bandpass Filters Tunable, See also Filters, Modem Filters	R5604	Reticon			TCM4913	TI			<b>MC145100</b>	<b>Motorola (3536)</b>		
	R5605	Reticon			TCM4914	TI			$\mu$ PD22100	NEC		
	R5606	Reticon			TCM4916	TI			<b>CD22100</b>	<b>RCA (840)</b>		
	R5609	Reticon			TCM4917	TI			M22100	SGS		
	R5611	Reticon			Codec/Filter, Serial Interface	TP3052	National		EFB7310	Thomson-CSF		
	R5612	Reticon			TP3053	National						
	R5613	Reticon			TP3054	National		Crosspoint Switch (4x4x2 with control memory)	<b>CD22101</b>	<b>RCA (840)</b>	145	
	R5614	Reticon			TP3057	National			<b>CD22102</b>	<b>RCA (840)</b>		
	R5615	Reticon			Codec/Filter, $\mu$ 255 Law	<b>S3507</b>	<b>AMI (3348)</b>		Crosspoint Switch (8x4 switch matrix)	MT8804B	† Mitel	
	R5616	Reticon			MT3507	Mitel						
	R5620	Reticon			MT3507A	Mitel		CTCSS (continuous tone-controlled squelch system)	MC1938	NEC	90	
	R5621	Reticon			MT8960	Mitel		Encoder/Decoder	MC4137	NEC		
	R5622	Reticon			MT8962	Mitel			MC4138	NEC		
Bandsplit Filter, DTMF	S3525A	AMI			MT8964	Mitel						
	S3525B	AMI			MK5300	Mostek		Current Limiter (for electromechanical telephone bells)	ZN474E	Ferranti	95	
	G8865X	GTEMicro			TCM2916	TI		Delta Modulation System, Continuously Variable Slope	XR3417	Exar	150	
	MV8865	Plessey			TCM4916	TI			XR3418	Exar		
	TT6177	Teltone			Codec/Filter, $\mu$ 255 Law PCM	S3501	AMI		XR3517	Exar		
Call Progress Tone Detector	M980	Teltone				S3502	AMI		XR3518	Exar		
	M981	Teltone			Codec, $\mu$ 255/A Law	ZNPCM1	Ferranti		<b>HC55536</b>	<b>Harris (3470)</b>	155	
	M982	Teltone				AY3-9900	GI		<b>HC55564</b>	<b>Harris (3469)</b>		
Click Suppressor	ZSY0	ITT				MC14404	Motorola		MC3417	Motorola		
Codec, A Law	5156	Fairchild				<b>MC14407</b>	<b>Motorola (3536)</b>		MC3418	Motorola		
	G8911	GTEMicro				TP5117A	National		MC3517	† Motorola		
	HC5156A	Harris				S291	Siemens		MC3518	† Motorola		
	HC5511	Harris				SM61C	Siemens					
	2911A	Intel			Codec, $\mu$ 255 Law	5116	Fairchild		Delta Sigma Modulator/Demodulator	ZNPCM2	Ferranti	160
	MK5156	Mostek				5151	Fairchild		Dialer, Microcomputer (clock, duration, fee pulse counter, redial, storage)	SAA6002	ITT	165
	TP3021	National				G8910	GTEMicro			MK5170	Mostek	
	TP5156	National				HC5116A	Harris			PCD3340	Signetics	
	TP5156A	National				HC5510	Harris		Dialer, Microcomputer (clock, stopwatch, redial, 16/32 number)	TZ2003	TI	
	M090	SGS				2910	Intel		Dialer, Loop Disconnect	DF820	ITT	
	M5156	SGS				2910A	Intel			DF821	ITT	
	S0291	Siemens				MK5116	Mostek			DF822	ITT	
	TCM2911A	TI				MK5151	Mostek					
	EFB7356	Thomson-CSF				MC14406	Motorola					
						TP3020	National					

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LINEAR Master Selection Guide

LINEAR-Telecommunication Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Telecommunication Circuits (Cont'd)</b>				<b>Telecommunication Circuits (Cont'd)</b>				<b>Telecommunication Circuits (Cont'd)</b>			
Dialer, Loop Disconnect				Digital Modulator, 1200/2400 bps	MC6172	Motorola (3536)	60	FSK Receiver, Wideband	MC3356	Motorola	
	DF823	ITT		Digital Time/Space Crosspoint Switch	MT8981	Mitel		Gyrator	AF120	National	110
	M760	SGS		Digital Trunk-Interface Module (T1)	MH89750	Mitel		HDB3/AMI Encoder/Decoder	TCM2202	TI	
	M760A	SGS		Echo Canceller Peripheral	S28216	AMI		HDB3/AMI Encoder/Decoder (0 to 2.5 MHz, T1 Compatible)	TCM2202	TI	
	PCD3324	Signetics	5	Elastic Store (provides buffer store and justification function in a 2nd order PCM multiplexer)	TCM2401	TI		HDB3/AMI Equipment Line Interface (interfaces signal, clock to line)	TCM2203	TI	
	PCD3325	Signetics		Electronic Phone Chip, Microprocessor Interface	MC34010	Motorola (3060)	65	HDB3/AMI Line Interface	TCM2203	TI	115
	EFB9158	Thomson-CSF		Electronic Telephone Circuit (DTMF), Tone Ringer, Speech Network and DC Line Voltage Regulator	MC34011	Motorola (3060)		HDB3 Transcoder	TCM2201	TI	
Dialer, Push Button Telephone Dialer Circuit (converts push button inputs to pulses)				Encoder 8 of 8 Keyboard to Binary Encoder (4x4 key input to 4 line output)	MC14419	Motorola (3536)		IR Receiver	SM802	Siemens	
	AY5-9171	GI		Filter Array, mask programmable switched capacitor	R5625	† Reticon		IR Transmitter	SM801	Siemens	
	MC14409	Motorola (3536)		Filter, Codec (A law/ $\mu$ law)	MB6001	Fujitsu	70	Limiter (limits voltage on two-wire speech)	U225	Telefunken	120
	MC34015	Motorola			MB6002	Fujitsu		Microphone Amplifier (for telephone circuits)	ZN470AE	Ferranti	
	MMS3143	National	10		MC14406	Motorola (3536)			ZN472E	Ferranti	
	MMS3144	National			MC14401	Motorola (3536)			ZN475E	Ferranti	
	MMS3190	National			MC14402	Motorola (3536)			ZN476E	Ferranti	
	MMS3393	National			MC14403	Motorola			ZN477E	Ferranti	125
	MMS3394	National			MC14405	Motorola		Modem Band-Pass Filter, 300 Baud	MC145440	Motorola	
	CRC8000	Rockwell	15	Filter, Low-Pass Programmable	S3528	AMI (3347)	75		MC145441	Motorola	
	CRC8001	Rockwell		Filter, PCM	AMS3064	Aptek		Modem Circuit, V.21, V.23, Bell 103, 202	AM7910	AMD (1433,3321)	
	M761	SGS			AMS3065	Aptek		Modem Circuit, V23 or Bell 202	EFB7910	Thomson-CSF	
	M761A	SGS			AF132	National		Modem Circuits, CCITT FSK V.23 or Bell 202	TCM3101	TI	130
	PCD3320	Signetics			AF133	National		Modem Circuits, for CCITT FSK System	CH1225	Cermetek	
	PCD3321	Signetics	20		AF134	National			CH1253	Cermetek	
	PCD3322	Signetics			MC4264	NEC			CH1258	Cermetek	
	PCD3323	Signetics			M5M8912	OKI (4118)			CH1259	Cermetek	
	TEA1021	Signetics			SM153	Siemens			CH1263	Cermetek	
	TEA1043	Signetics			TCM2912A	TI			CH1268	Cermetek	
	TEA1044	Signetics	25	Filter, PCM Transmit/Receive	G8912B	GTEMicro		Modem Circuits, for 103/113 and Other FSK Systems	EFB7510	Thomson-CSF	
	DF320	Siliconix			HC5512	Harris (3468,3468)			CH1222	Cermetek	
	DF320A	Siliconix			HC5512A	Harris (3468,3468)			CH1223	Cermetek	
	DF322	Siliconix			HC5512C	Harris (3468)			CH1224	Cermetek	
	DF326	Siliconix			2912A	Intel			CH1252	Cermetek	
	TCM1101	TI	30		MK5912	Mostek			CH1257	Cermetek	
	EFB9151	Thomson-CSF			MK8912	Mostek			CH1262	Cermetek	
	EFB91E51	Thomson-CSF			MC14413-1	Motorola (3536)	45		CH1265	Cermetek	
Dialer, Push Button Telephone Dialer Circuit (converts push button inputs to pulses)					MC14413-2	Motorola (3536)			CH1266	Cermetek	145
	S2580A	AMI (3341)			MC14414-1	Motorola (3536)			CH1267	Cermetek	
	AY5-9151	GI			MC14414-2	Motorola (3536)			CH1267	Cermetek	
	AY5-9153	GI			TP3040	National			CH1271	Cermetek	
	AY5-9154	GI			TP3040A	National			CH1272	Cermetek	
	DF320	ITT	35		CD22413	RCA (840)			CH1273	Cermetek	
	MT4320	Mitel			CD22414	RCA (840)			CH1280	Cermetek	150
	MT4322	Mitel			M5912	SGS			AY7-1203	GI	
	MT4323	Mitel			TCM2912B	TI			MC145445	Motorola (3536)	
	MT4325	Mitel			EFB7912	Thomson-CSF			MC145450	Motorola (3536)	
	MT4326	Mitel			Filter, Speech (dual, tunable low-pass)	MC145414	Motorola (3536)		MM54HC942	† National	
	MT4327	Mitel			Filter, Tunable (lowpass/bandpass)	MC145431	Motorola	105	MM74HC942	National	155
	MK50981	Mostek			Filter, Tunable (notch, bandpass)	MC145433	Motorola (3536)		Modem Circuits, FSK CCITT V.21 or Bell 103	TMS99532	TI (1712)
	MK50982	Mostek			MC145434	Motorola			TMS99534	TI (1712)	
	MK50991	Mostek			Filter, Dual Low Pass Linear Phase	MC145415	Motorola (3536)		Modem Circuits, PSK CCITT V.22, V.26 or Bell 212A, 201BC	XR2123	Exar (3370)
	MK50992	Mostek							Modem Circuits, PSK Demodulator, Bell 212A	XR2122	Exar (3370)
	MK50992	Mostek									
	MK50992	Mostek									
	MC14498	Motorola (3536)									
Dialer, Repertory											
	S25610	AMI (3342)									
	S25610E	AMI (3342)									
	S2562	AMI									
	S2563	AMI									
	CET200	Fairchild									
	MK5175	Mostek									
	SFF19200	Thomson-CSF									
Digital Filter with Utility Peripheral	S28215	AMI									
Digital Line Interface Controllers											
	TP3110	National									
	TP3120	National									
Digital Line Interface Module											
	MH89700	Mitel									

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Telecommunication Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Telecommunication Circuits (Cont'd)</b>				<b>PCM Signal Monitor</b>				<b>Telephone Chipset (microcomputer-compatible)</b>			
Modem Circuits, PSK Modulator, Bell 212A				SP 1450B Plessey				TCM1707 TI			
XR2121 Exar (3369)				SP 1455B Plessey				TCM1708 TI			
Modem Circuits, Sync/Async Converter				<b>Polynomial Generator/Checker (for character-oriented data communication links)</b>				<b>Telephone Restrictor</b>			
XR2125 Exar (3371)				SCN2853A Signetics (1671)				CRN3300 Thomson-CSF			
Modem Filter digitally programmable transversal filter				RC Encoder, Programmable 7-Channel				Telephone Speech Circuit (programmable)			
R5910 Reticon				NE5044 Signetics (3626)				LS288 SGS			
Modem Filter, FSK				RD Decoder, Seven Channel				Teletex Acquisition/Control			
XR2103 Exar (3368)				NE5045 Signetics (3629)				SAA5040B Signetics			
Modem Filter, General Purpose (bandpass filter and MUX)				Receiver, DTMF (hybrid)				Teletex Adaptive Data Slicer (provides data, clock and synchronizing signals)			
R5633 Reticon				MH88315 Mitel				SN76940 TI			
Modem Filter, PSK				Relay Driver (for 48 V telephone relays)				Teletex Character Generator, American			
XR2120 Exar (3368)				DS1686 † National				SAA5055 Signetics			
Modem Filter with Equalizer, 212A/V.22				DS1687 † National				Teletex Timing Circuit, 525 Lines			
S3522 AMI (3345)				DS3686 National				SAA5025B Signetics			
Modem Filter (300 baud)				DS3687 National				SAA5025C Signetics			
MC145440 Motorola (3536)				UDN-2580A Sprague				SAA5025D Signetics			
MC145441 Motorola (3536)				UDN-2585A Sprague				Teletex Timing Circuit, 625 Lines			
MC6170 Motorola				UDN-2588A Sprague				SAA5020 Signetics			
R5630 Reticon				UDN-2956A Sprague				Teletex Video Input Processor			
R5631 Reticon				UDN-2957A Sprague				SAA5030 Signetics			
Modem Filter (1200 baud)				Relay Driver, Quad Negative Voltage				Teletext GALA			
CH1710 Cermetek				DS3680 National				SAA5045 Signetics			
CH1720 Cermetek				DS3680 TI				Time Division Multiplexing, Demultiplexor			
CH1730 Cermetek				Ring Detector (operates from 16 to 68 Hz)				EFB7334 Thomson-CSF			
CH1740 Cermetek				TCM1520A TI				Time Division Multiplexing, Serial/Parallel Shift/Conversion			
Modem Filter, 1200 Baud, 212/V.22				Serial Receiver (changes T1-D2 or T1-D3 input to parallel output)				EFB7303 Thomson-CSF			
R5632 Reticon				R8060 Rockwell				Time Division Multiplexing, Temporal Matrix for 8x8 frame switching			
Modem Filter, (1200/150 Baud)				Serial Transmitter (generates 193-Bit data stream in T1-D2 or T1-D3 format)				EF7331 Thomson-CSF			
R5634 Reticon				R8050 Rockwell				Time Division Multiplexing, 2.048 MHz Clock Recovery			
Modem Filter (1200/2400 baud)				Signal Processing Peripheral (arithmetic processor, RAM, ROM, I/O)				EF7332 Thomson-CSF			
MC3862 Motorola (1505)				S28211A AMI (3348)				Time Division Multiplexing, 2.048 MHz Communication Terminal			
Modem (for serial communications at up to 600 b/s)				S28211B AMI (3348)				EF7332 Thomson-CSF			
MC3866 Motorola (1686, 3536)				Speech Circuit for Electronic Telephones				Time Slot Access Circuit			
Modem, FSK Demodulator				TP5700 National				MJ1446 Plessey			
XR2211 † Exar				TP5710 National				Time Slot Assigner Circuit			
XR2211 Exar				PBL3725 RIFA				MC14416 Motorola (3536)			
XR2211M † Exar				PBL3726 RIFA				MC14417 Motorola (3536)			
Modem, FSK Modulator				TCM1705A TI				MC14418 Motorola (3536)			
XR2206 † Exar				Subscriber Line Interface Circuit (hybrid SIP)				EFB7336 Thomson-CSF			
XR2206M † Exar				MH88505 Mitel				Tone Caller			
XR2207 Exar				MH88506 Mitel				ZN473E Ferranti			
XR2207M Exar				MH88510 Mitel				Tone Decoder (operates from 0.01 Hz to 500 kHz)			
Modem Interface and Receiver Filter Circuit (300 baud modem)				S2500A AMI				XR2211C Exar			
CH1230 Cermetek				S2550A AMI (3344)				XR2211M † Exar			
Modem System, FSK				AMS2001 Aptek				XR567AC Exar			
XR14412F Exar (3368)				AMS2002 Aptek				XR567AM † Exar			
XR14412Y Exar (3369)				AMS2006 Aptek				LM567C National			
MC14412 Motorola (3536)				AMS3081 Aptek				XR2211C Raytheon			
Modem System, FSK, 1200 Baud				AMS3082 Aptek				XR2211M † Raytheon			
S8180 SiliconSys (3665)				AMS3084 Aptek				NE567 Signetics			
Modem, 2400 bps Demodulator				AMS3085 Aptek				SE567 † Signetics			
MC8178 Motorola (3536)				HC5502 Harris (3486)				Tone Decoder, Dual (operates from 0.01 Hz to 500 kHz)			
Multifrequency Interface Line Circuit				HC5504 Harris (3467)				XR2567 Exar (3367)			
LS342 SGS				MH88500 Mitel				XR2567M † Exar			
Multifrequency Tone Generator (generates 15 tone pairs)				MC34019 Motorola				Tone Decoder, Dual			
AMS3501 Aptek				MC3418 Motorola (3060)				MT8860 Mitel			
Multifrequency Tone Generator (module)				MC3419A Motorola (3060)				MT8862 Mitel			
CM7050 Mitel				MC3419C Motorola (3060)				MT8863 Mitel			
PCM Repeater				MC3519 Motorola				MV8860 Plessey			
XRC240 Exar				LS285A SGS				MV8862 Plessey			
XRC262 Exar				G150 Siemens				MV8863 Plessey			
XRC277 Exar				ST120 Signetics				CRC8030 Rockwell			
TCM2101 TI				TCM4204 TI				SS1202 SiliconSys			
TCM2102 TI				TCM4205 TI				Tone Receiver Circuit, Dual (combine with filters for touch tone receiver)			
				TCM4207 TI				G8860 GTEMicro			
								G8860X GTEMicro			
								MK5102-5 Mostek			
								(Continued)			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Telecommunication Circuits (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line				
<b>Telecommunication Circuits (Cont'd)</b>				<b>Dual Tone Generator (touchtone frequencies)</b>							
Tone Receiver Circuit, Dual (combine with filters for touch tone receiver) (Cont'd) MK5103-5 Mostek SM301A Siemens M967 Teltone TT6174 Teltone <hr/> Tone Receiver Circuit, Dual (with filters) AMS3201 Aptek AMS3210 Aptek AMS88205-5NC Aptek G8870 GTEMicro MH88210 Mitel MH88305 Mitel MT8870 Mitel SSI201 SiliconSys (3685) SSI202 SiliconSys (3685) M927 Teltone M937 Teltone M947 Teltone M948 Teltone M956 Teltone M957 Teltone <hr/> Tone Ringer, (operates from ring signal, drives piezo element) <b>MC34012 Motorola (3060)</b> Tone Ringer (operates from ring signal, drives speaker) S2581A AMI (3343) S2581C AMI (3343) CS8204 Cherry CS8205 Cherry ML8204 Mitel ML8205 Mitel MC34012-1 Motorola (3060) MC34012-2 Motorola (3060) MC34012-3 Motorola (3060) M764 SGS M764A SGS TCM1501A TI TCM1506A TI TCM1512A TI TCM1513A TI TCM1520A TI <hr/> Transcoder, with Error Detection (10 kb/s to 2.048 Mb/s) <b>CD22103 RCA (840)</b> Transmission Interface TEA1042 Signetics TEA1046 Signetics TEA1055 Signetics <hr/> Tri-Port Memory (T1 carrier) R8040 Rockwell Universal Digital Loop Transceiver, Four Wire MC145420 Motorola Universal Digital Loop Transceiver, Four Wire Slave MC145425 Motorola Universal Digital-Loop Transceiver MC145422 Motorola (3536) MC145423 Motorola MC145426 Motorola (3536) <hr/> Videotex Decoder, Interactive SAA5070 Signetics Single Channel Signaling Controller EFB7335 Thomson-CSF Single Frequency Tunable Bandpass/Notch Filter/Tone Generator <b>S3526 AMI (3346)</b> Dual Tone Generator (generates 12-16 tone pairs) AY3-9400 GI AY3-9410 GI				S25089 AMI (3340) S2559A AMI (3339) S2559B AMI (3339) S2559C AMI (3339) S2559D AMI (3339) S2559E AMI (3339) S2559F AMI (3339) S2559G AMI (3339) S2559H AMI (3339) S2859 AMI S2860 AMI AY5-9480 GI AY5-9559 GI ICM7206A Intersil (3524) ICM7206B Intersil (3524) ICM7206C Intersil (3524) ICM7206D Intersil (3524) SBA5089 ITT SBA5091 ITT MT5089 Mitel MK5087 Mostek MK5089 Mostek MK5091 Mostek MK5092 Mostek MK5094 Mostek MK5380 Mostek MK5382 Mostek MK5389 Mostek MC14410 Motorola (3536) MM53125 National MM53130 National MM5395 National μPC767 NEC CD22859 RCA (840) PBD3534 RIFA PBD3535 RIFA M751 SGS S359 Siemens PCD3311 Signetics PCD3312 Signetics TCM5087 TI TCM5089 TI TCM5091 TI TCM5092 TI EFB7189 Thomson-CSF							
				<hr/> Tone Ringer, (operates from ring signal, drives piezo element) <b>MC34012 Motorola (3060)</b> Tone Ringer (operates from ring signal, drives speaker) S2581A AMI (3343) S2581C AMI (3343) CS8204 Cherry CS8205 Cherry ML8204 Mitel ML8205 Mitel MC34012-1 Motorola (3060) MC34012-2 Motorola (3060) MC34012-3 Motorola (3060) M764 SGS M764A SGS TCM1501A TI TCM1506A TI TCM1512A TI TCM1513A TI TCM1520A TI <hr/> Transcoder, with Error Detection (10 kb/s to 2.048 Mb/s) <b>CD22103 RCA (840)</b> Transmission Interface TEA1042 Signetics TEA1046 Signetics TEA1055 Signetics <hr/> Tri-Port Memory (T1 carrier) R8040 Rockwell Universal Digital Loop Transceiver, Four Wire MC145420 Motorola Universal Digital Loop Transceiver, Four Wire Slave MC145425 Motorola Universal Digital-Loop Transceiver MC145422 Motorola (3536) MC145423 Motorola MC145426 Motorola (3536) <hr/> Videotex Decoder, Interactive SAA5070 Signetics Single Channel Signaling Controller EFB7335 Thomson-CSF Single Frequency Tunable Bandpass/Notch Filter/Tone Generator <b>S3526 AMI (3346)</b> Dual Tone Generator (generates 12-16 tone pairs) AY3-9400 GI AY3-9410 GI				S2569 AMI S2569A AMI <hr/> Two-Wire-to-Four Wire Converter (replaces hybrid transformer) LS285A SGS ST120 Signetics			
								S2569 AMI S2569A AMI <hr/> Two-Wire-to-Four Wire Converter (replaces hybrid transformer) LS285A SGS ST120 Signetics			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR

Master Selection Guide

LINEAR-Timers

Function	Device	Source	Line	Function	Device	Source	Line
<b>Timers</b>				Timer (CMOS 555), Low Power			
	<b>ICM7555</b>	<b>Intersil</b>	<b>(3527)</b>		<b>ICM7555M</b>	<b>† Intersil</b>	<b>(3527)</b>
Adjustable Threshold	MC1422	Motorola		Timer with + 12 Counter (time constant 7200 RC)	ZN1034E	Ferranti	60
Delayed Switch ON/OFF (for driving relays, lamps, etc.)	PBA3008	RIFA		Timer with Counter, (time constant 128 RC)	XR2242C	Exar	
	PBA3009	RIFA			XR2242M	† Exar	
Internal Current Source	XR320	Exar		Universal (includes two 4-digit counters, comparator, clock oscillator and divider)	MM5865	National	5
Long Period (1 sec to 1 week)	AN6780	Panasonic		Dual (CMOS 556) Low Power	<b>ICM7556</b>	<b>Intersil</b>	<b>(3527)</b>
Long Period (10 ms to 10,000 hours)	MW122	AnalogSys			<b>ICM7556M</b>	<b>† Intersil</b>	<b>(3527)</b>
Low Power (supply 3 V)	SN76812	TI		Dual (dual 555), Pinout Variations	XR2556C	Exar	65
Micropower, Long Range (microseconds to days)	XR2243	Exar			XR2556M	† Exar	
Microwave Oven Controller	TMS1117	TI		Dual (dual 555)	XR556C	Exar	
	TMS1117NLP	TI	10		XR556M	† Exar	
Precision	MW107	AnalogSys			XRL556C	Exar	70
	MW197	AnalogSys			XRL556M	† Exar	
	LM122	† National			μA556C	Fairchild	
	LM2905	National			NE556	Intersil	
	LM322	National			SE556	† Intersil	
	LM3905	National	15		MC3456	Motorola	75
Precision Decade, Direct 7 Segment and Digit Drivers	ICM7045A	Intersil			MC3556	† Motorola	
Programmable (Includes a counter for long time delays)					LM556	† National	
Binary Control					LM556C	National	
	XR2240C	Exar			RC556	Raytheon	
	XR2240M	† Exar			RM556	† Raytheon	80
	μA2240C	Fairchild			NE556	Signetics	
	μA2240M	† Fairchild	20		NE556-1	Signetics	
	ICM7240	Intersil			SA556	Signetics	
	ICM7242	Intersil			SA556-1	Signetics	
	ICM7250	Intersil			SE556	† Signetics (526)	85
	ICM7260	Intersil			SE556-1	† Signetics (526)	
	LS7210	LSIComp	(803)		SE556-1C	† Signetics (526)	
	MC14541BA	† Motorola			SG556	† SiliconG	
	MC14541BC	Motorola			SG556C	SiliconG	
	μA2240C	TI			NE556	TI	90
	μA2240M	† TI			SE556	† TI	
Timer	XR555C	Exar		Quad, Current Sink Output	XR558C	Exar	
	XR555M	† Exar			XR558M	† Exar	
	XRL555	Exar			NE558	Signetics (3670)	95
	μA555C	Fairchild			SA558	Signetics (3670)	
	NE555	Intersil			SE558	† Signetics (3670)	
	SE555	† Intersil	35	Quad, Current Source Output	XR559C	Exar	
	MC1455	Motorola			XR559M	† Exar	
	MC1555	† Motorola					
	LM555C	National					
	μPC1555	NEC	40				
	RC555	Raytheon					
	RM555	† Raytheon					
	CA555	† RCA (3804)					
	CA555C	RCA (3804)					
	NE555	Signetics	45				
	SE555	† Signetics (526)					
	SE555C	† Signetics (526)					
	SG555	† SiliconG					
	SG555C	SiliconG					
	ULN-2430M	Sprague	50				
	355A/C	TeledyneS					
	355B/M	† TeledyneS					
	NE555	TI					
	SE555	† TI					
	TD0555	Thomson-CSF	55				
	TA7326	Toshiba					
	TA7327	Toshiba					

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR Master Selection Guide

LINEAR-Voltage Regulators

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line
<b>Fixed, Positive</b>					5	1000			(Cont'd)
For low amounts of current, voltage references can be used as voltage regulators. See Linear—Other Devices, Reference Diodes									
2.6	100	MC78L02C	Motorola				L7805	† SGS	
		μA78L02AC	TI				L7805C	SGS	
		μA78L02C	TI				SG109K	† SiliconG	70
		TA7316	Toshiba	5			SG209K	SiliconG	
							SG309K	SiliconG	
							SG7805ACP	SiliconG	
							SG7805AP	† SiliconG	
							SG7805CP	† SiliconG	
							SG7805P	SiliconG	75
5	100	μA78L05C	Fairchild				SFC2109M	Thomson-CSF	
		LM2931-5	Motorola				SFC2805	† Thomson-CSF	
		MC78L05AC	Motorola				TA78005	Toshiba	
		MC78L05C	Motorola				<b>UC7805A</b>	<b>Unitrade (3711)</b>	80
		LM140LA-5	National	10			<b>UC7805AC</b>	<b>Unitrade (3711)</b>	
		LM78L05A	National			1500	μA7805C	Fairchild	
		LM78L05C	National				μA7805M	† Fairchild	
		μPC78L05	NEC				LAS1505	Lambda	
		μA78L05AC	TI				LAS15A05	Lambda	
		μA78L05C	TI	15			<b>LM140-5</b>	<b>† Motorola (3531)</b>	85
		TA78L005	Toshiba				<b>LM340-5</b>	<b>Motorola (3531)</b>	
		TA78L005A	Toshiba				<b>MC7805</b>	<b>† Motorola (3531)</b>	
	150	LM3931-5	National				<b>MC7805A</b>	<b>† Motorola (3531)</b>	
		LM2930-5	TI				<b>MC7805AC</b>	<b>Motorola (3531)</b>	
		LM330-5	TI	20			<b>MC7805C</b>	<b>Motorola (3531)</b>	90
	200	<b>LM109H</b>	<b>Motorola (3531)</b>				LM140-5	† National	
		<b>LM209H</b>	<b>Motorola (3531)</b>				LM140A-5	† National	
		<b>LM309H</b>	<b>Motorola (3531)</b>				LM2931-5	National	
		LM109H	† National				LM340-5	National	
		LM209H	National	25			LM340A-5	National	95
		LM2930-5	National				LM7805	National	
		LM309H	National				SG140-05K	SiliconG	
		LM330-5	National				SG140-05R	SiliconG	
		SG109T	† SiliconG				SG340-05K	SiliconG	
		SG209T	SiliconG	30			SG340-05R	SiliconG	100
		SG309T	SiliconG				SG7805ACK	SiliconG	
		SFC2109	† Thomson-CSF				SG7805ACR	SiliconG	
		SFC2209	Thomson-CSF				SG7805AK	† SiliconG	
		SFC2309	Thomson-CSF				SG7805CK	SiliconG	
	250	LM342-5	National	35			SG7805CR	SiliconG	105
	400	L4805	SGS				SG7805K	† SiliconG	
	500	μA78M05C	Fairchild				SG78R05	† SiliconG	
		μA78M05M	† Fairchild				LM340-5	TI	
		TDD1605	ITT				TL780-05C	TI	
		MC7705C	Motorola	40			<b>UC7805</b>	<b>† Unitrade (3711)</b>	110
		MC78M05C	Motorola				<b>UC7805C</b>	<b>Unitrade (3711)</b>	
		LM341-5	National			2000	HA17805P	Hitachi	
		LM78M05	National				LAS1605	Lambda	
		μPC78M05	NEC				LAS1605B	Lambda	
		L194-5	SGS	45			L78S05	† SGS	115
		L2605	SGS				L78S05C	SGS	
		L4705	SGS						
		L487	SGS						
		SG140-05T	† SiliconG			3000	SH123	† Fairchild	
		SG340-05T	SiliconG	50			SH223	Fairchild	
		SG7805ACT	SiliconG				SH323	Fairchild	
		SG7805AT	† SiliconG				μA123	Fairchild	
		μA78M05C	TI				μA223	† Fairchild	
		μA78M05M	† TI				μA323	Fairchild	
							LAS1405	Lambda	120
							LAS1405B	Lambda	
	750	LM2935	National	55			LM123	† LinearTech	125
	850	L129	SGS				LM123A	† LinearTech	
	1000	μA109M	† Fairchild				LM323	LinearTech	
		μA209M	Fairchild				LM323A	† LinearTech	
		μA309C	Fairchild				<b>LM123</b>	<b>† Motorola (3531)</b>	130
		<b>LM109K</b>	<b>† Motorola (3531)</b>	60			<b>LM123A</b>	<b>† Motorola (3531)</b>	
		<b>LM209K</b>	<b>Motorola (3531)</b>				<b>LM223</b>	<b>Motorola (3531)</b>	
		<b>LM309K</b>	<b>Motorola (3531)</b>				<b>LM223A</b>	<b>Motorola (3531)</b>	
		LM109K	† National				<b>LM323</b>	<b>Motorola (3531)</b>	
		LM209K	National				<b>LM323A</b>	<b>Motorola (3531)</b>	
		LM309K	National	65			MC78T05	Motorola	135
		μPC7805	NEC				MC78T05A	† Motorola	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Voltage Regulators (Cont'd)

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line		
<b>Fixed, Positive (Cont'd)</b>					<b>(Cont'd)</b>						
5	3000	MC78T05AC	Motorola	5	6	2000	HA17806P	Hitachi	70		
		MC78T05C	Motorola		CJSE017		Solitron				
		LM123	† National		CJSE019		Solitron				
		LM223	National		3000	CJSE021	Solitron				
		LM323	National			LAS1406	Lambda				
		SG123	† SiliconG		MC78T06	† Motorola					
		SG153-05	† SiliconG		MC78T06C	Motorola					
		SG223	SiliconG		10000	42050-610	Micropac				
	SG253-05	SiliconG	42055-620	Micropac							
	5000	5000	μA78H05	Fairchild	10	6.2	100	μA78L62C	Fairchild	80	
			μA78H05A	Fairchild		μA78L06AC		TI			
			LAS1905	Lambda		μA78L06C		TI			
LAS1905B			Lambda	100		TA78L007	Toshiba				
LT1003C			LinearTech			TA78L007A	Toshiba				
LT1003M			† LinearTech	10000		42050-710	Micropac				
42050-055			Micropac			42055-720	Micropac				
MLM138-05			Micropac	100		TA78L075	Toshiba				
MLM338-05			Micropac			TA78L075A	Toshiba				
8000			8000	LAS3905		Lambda	20	1000	L7875	† SGS	85
				LAS3905K		Lambda			L7875C	SGS	
10000			10000	μA78P05		Fairchild	25	100	L78C75C	SGS	90
	42050-510	Micropac		L78S75	† SGS						
20000	20000	LML196-05	Micropac	30	150	MC78L08AC	Motorola	95			
		MLM396-05	Micropac			MC78L08C	Motorola				
20000	20000	MPC100	Motorola	35	200	μPC78L08	NEC	100			
		42055-520	Micropac			μA78L08AC	TI				
6	100	TA78L006	Toshiba	40	500	μA78L08C	TI	95			
		TA78L006A	Toshiba			TA78L008	Toshiba				
	500	500	μA78M06C		Fairchild	45	150	TA78L008A	Toshiba	95	
			μA78M06M		† Fairchild			LM2930-8	TI		
			TDD1606		ITT		200	LM2930-8	National		
			MC78M06C		Motorola			500	μA78M08C	Fairchild	
			SG140-06T		† SiliconG		μA78M08M		† Fairchild		
			SG340-06T		SiliconG		TDD1608	ITT			
			SG7806ACT		SiliconG		MC78M08C	Motorola			
			SG7806AT		† SiliconG		μPC78M08	NEC			
	SG7806CT	SiliconG	SG140-08T		† SiliconG						
	SG7806T	† SiliconG	SG340-08T		SiliconG						
μA78M06C	TI	SG7808ACT	SiliconG								
μA78M06M	† TI	SG7808AT	† SiliconG								
1000	1000	SG7806ACP	SiliconG	50	1500	SG7808CT	SiliconG	110			
		SG7806AP	† SiliconG			SG7808T	† SiliconG				
		SG7806CP	SiliconG			μA78M08C	TI				
		SG7806P	† SiliconG			μA78M08M	† TI				
1500	1500	SFC2806	Thomson-CSF	60	1500	SFC2808	Thomson-CSF	115			
		μA7806C	Fairchild			μA7808C	Fairchild				
		LAS1506	Lambda		μA7808M	† Fairchild					
		LM140-6	† Motorola (3531)		HA17808	Hitachi					
		LM340-6	Motorola (3531)		LAS1508	Lambda					
		MC7806	† Motorola (3531)		LM140-8	† Motorola (3531)					
		MC7806A	† Motorola (3531)		LM340-8	Motorola (3531)					
		MC7806AC	Motorola (3531)		MC7808	† Motorola (3531)					
		MC7806C	Motorola (3531)		MC7808A	† Motorola (3531)					
		SG140-06K	SiliconG		MC7808AC	Motorola (3531)					
		SG140-06R	SiliconG		MC7808C	Motorola (3531)					
		SG340-06K	SiliconG		SG140-08K	SiliconG					
SG340-06R	SiliconG	SG140-08R	SiliconG								
SG7806ACK	SiliconG	SG340-08K	SiliconG								
SG7806ACR	SiliconG	SG340-08R	SiliconG								
SG7806AK	† SiliconG	SG7808ACK	SiliconG								
SG7806CK	SiliconG	SG7808ACR	SiliconG								
SG7806CR	SiliconG	SG7808AK	† SiliconG								
SG7806K	† SiliconG	SG7808CK	SiliconG								
SG78R06	† SiliconG	SG7808CR	SiliconG								
μA7806C	TI	SG7808K	† SiliconG								
		SG78R08	SiliconG								
		μA7808C	TI								

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR Master Selection Guide

LINEAR-Voltage Regulators (Cont'd)

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line
<b>Fixed, Positive</b>					<b>(Cont'd)</b>				
8	2000	LAS1608	Lambda		12	500	L194-12	SGS	65
	3000	MC78T08	Motorola				SG140-12T	† SiliconG	
		MC78T08C	Motorola				SG340-12T	SiliconG	
		SG153-08	† SiliconG				SG7812ACT	SiliconG	
		SG253-08	SiliconG	5			SG7812AT	† SiliconG	70
		SG353-08	SiliconG				SG7812CT	† SiliconG	
	5000	μA78H08C	Fairchild				SG7812T	† SiliconG	
	10000	μPC7808	NEC				μA78M12C	TI	
		SG7808ACP	SiliconG			720	μA78M12M	† TI	
		SG7808AP	† SiliconG	10			L130	SGS	
		SG7808CP	SiliconG			1000	μPC7812	NEC	75
		SG7808P	† SiliconG				STK521	Sanyo	
	20000	42055-820	Micropac				STK541	Sanyo	
8.2	100	μA78L82C	Fairchild				SG7812ACR	SiliconG	
8.5	400	L4885	SGS	15			SG7812AR	† SiliconG	
	500	L2685	SGS				SG7812CR	SiliconG	80
		L4785	SGS				SG7812R	† SiliconG	
	1000	μA7885	Fairchild				SFC2812	Thomson-CSF	
		μA7885C	TI				TA78012	Toshiba	
9	100	μA78L09C	Fairchild	20			<b>UC7812</b>	<b>† Unintred (3711)</b>	
		μA78L09AC	TI				<b>UC7812C</b>	<b>Unintred (3711)</b>	85
		μA78L09C	TI			1500	μA7812C	Fairchild	
		TA78L009	Toshiba				μA7812M	† Fairchild	
		TA78L009A	Toshiba				LAS1512	Lambda	
	1500	L7809	SGS	25			LAS15A12	Lambda	
		L78S09	† SGS				<b>LM140-12</b>	<b>† Motorola (3531)</b>	90
		SL78S09C	SGS				<b>LM340-12</b>	<b>Motorola (3531)</b>	
	20000	42055-920	Micropac				<b>MC7812</b>	<b>† Motorola (3531)</b>	
10	100	μA78L10AC	TI	30			<b>MC7812A</b>	<b>† Motorola (3531)</b>	
		μA78L10C	TI				<b>MC7812AC</b>	<b>Motorola (3531)</b>	
		TA78L010	Toshiba				<b>MC7812C</b>	<b>Motorola (3531)</b>	95
		TA78L010A	Toshiba				LM140-12	† National	
	200	LH0075	† National				LM140A-12	† National	
	400	L4810	SGS				LM340-12	National	
	500	TDD1610	ITT	35			LM340A-12	National	
		L2610	SGS				LM7812	National	100
		L4710	SGS				L7812	† SGS	
		μA78M10C	TI				L7812C	SGS	
	1500	LAS1510	Lambda	40			SG140-12K	† SiliconG	
		μA7810C	TI				SG140-12R	† SiliconG	
	2000	L78S10	† SGS				SG340-12K	SiliconG	105
		L78S10C	SGS				SG340-12P	SiliconG	
	9000	42050-109	Micropac				SG340-12R	SiliconG	
	20000	42055-1020	Micropac				SG7812ACK	SiliconG	
12	100	μA78L12C	Fairchild	45			SG7812ACP	SiliconG	
		MC78L12AC	† Motorola				SG7812AK	† SiliconG	110
		MC78L12C	Motorola				SG7812CK	SiliconG	
		LM140LA-12	† National				SG7812CP	SiliconG	
		LM340LA-12	National				SG7812K	† SiliconG	
		LM78L12A	National	50			SG78R12	† SiliconG	
		LM78L12C	National				LM340-12	TI	115
		μPC78L12	NEC				TL780-12C	TI	
		μA78L12AC	TI				μA7812C	TI	
		μA78L12C	TI				<b>UC7812A</b>	<b>† Unintred (3711)</b>	
		TA78L012	Toshiba	55			<b>UC7812AC</b>	<b>Unintred (3711)</b>	
		TA78L012A	Toshiba			2000	HA17812P	Hitachi	120
	250	LM342-12	National	45			LAS1612	Lambda	
	500	μA78M12C	Fairchild				LAS1612B	Lambda	
		μA78M12M	† Fairchild				STK531	Sanyo	
		TDD1612	ITT	60			L7812	† SGS	
		MC78M12C	Motorola				L7812C	SGS	125
		LM341-12	National				L78S12C	SGS	
		LM78M12	National			3000	LAS1412	Lambda	
		μPC78M12	NEC				LAS1412B	Lambda	
							MC78T12	† Motorola	
							MC78T12A	† Motorola	
							MC78T12AC	Motorola	
							MC78T12C	Motorola	
							SG153-12	† SiliconG	130

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



LINEAR-Voltage Regulators (Cont'd)

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line	
<b>Fixed, Positive</b>					<b>(Cont'd)</b>					
12	3000	SG253-12	SiliconG	5	15	1500	MC7815A	† Motorola (3531)	70	
		SG353-12	SiliconG					MC7815AC		Motorola (3531)
	5000	μA78H12A	Fairchild					MC7815C		Motorola (3531)
		LAS1912	Lambda					LM140-15		† National
		LAS1912B	Lambda					LM140A-15		† National
		MLM138-12	Micropac					LM340-15		National
		MLM338-12	Micropac					LM7815		National
	8000	42050-128	Micropac					SG140-15K		† SiliconG
	10000	MLM196-12	Micropac					SG140-15R		† SiliconG
		MLM396-12	Micropac					SG7815ACK		SiliconG
	16000	42055-1216	Micropac				SG7815ACP	SiliconG		
13.2	100	TA78L132	Toshiba	10			SG7815AK	† SiliconG	75	
		TA78L132A	Toshiba					SG7815CK		SiliconG
	1500	LAS15CB	Lambda				SG7815CP	SiliconG	80	
13.8	2000	LAS16CB	Lambda				SG7815K	† SiliconG		
14	2000	LAS16CB	Lambda	15			SG78R15	SiliconG	85	
	8000	42050-148	Micropac					LM340-15		TI
	16000	42055-1416	Micropac					TL780-15C		TI
15	100	μA78L15C	Fairchild	20		2000	HA17815P	Hitachi	90	
		MC78L15AC	Motorola					LAS1615		Lambda
		MC78L15C	Motorola					LAS1615B		Lambda
		LM140LA-15	† National					L78S15		† SGS
		LM340LA-15	National					L78S15C		SGS
		LM78L15A	National					CJSE001		Solitron
		LM78L15C	National					CJSE003		Solitron
		μPC78L15	NEC					CJSE005		Solitron
		μA78L15AC	TI				3000	LAS1415		Lambda
		μA78L15C	TI					LAS1415B		Lambda
		TA78L015	Toshiba				MC78T15	† Motorola		
		TA78L015A	Toshiba				MC78T15A	† Motorola		
	250	LM342-15	National	25			MC78T15AC	Motorola	95	
	500	μA78M15C	Fairchild					MC78T15C		Motorola
		μA78M15M	† Fairchild				SG153-15	† SiliconG		
		TDD1615	ITT				SG253-15	SiliconG		
		MC78M15C	Motorola				SG353-15	SiliconG		
		LM341-15	National			5000	LAS1915	Lambda		
		LM78M15	National				LAS1915B	Lambda		
		μPC78M15	NEC				LML138-15	Micropac		
		L194-15	SGS				LML338-15	Micropac		
		SG140-15T	† SiliconG			8000	42050-158	Micropac		
		SG7815ACT	SiliconG				MLM196-15	Micropac		
		SG7815AT	† SiliconG				MLM396-15	Micropac		
		SG7815CT	SiliconG				42055-1516	Micropac		
		SG7815T	† SiliconG		16	8000	42050-168	Micropac		
		μA78M15C	TI			12000	42055-1612	Micropac		
		μA78M15M	† TI		18	100	MC78L18AC	Motorola		
	600	L131	SGS	40			MC78L18C	Motorola		
	1000	L7815	† SGS					TA78L018	Toshiba	
		L7815C	SGS				TA78L018A	Toshiba		
		SG7815ACR	SiliconG			500	TDD1618	ITT		
		SG7815AR	† SiliconG				MC78M18C	Motorola		
		SG7815CR	SiliconG				μPC78M18	NEC		
		SG7815R	† SiliconG				SG140-18T	† SiliconG		
		TA78015	Toshiba				SG340-18T	SiliconG		
		UC7815	† Unitrode (3711)				SG7818ACT	SiliconG		
		UC7815A	† Unitrode (3711)				SG7818AT	† SiliconG		
		UC7815AC	Unitrode (3711)				SG7818CT	SiliconG		
		UC7815C	Unitrode (3711)				SG7818T	† SiliconG		
	1500	μA7815C	Fairchild	50		1000	μPC7818	NEC	125	
		μA7815M	† Fairchild					STK522		Sanyo
		LAS1515	Lambda				STK542	Sanyo		
		LAS15A15	Lambda				L7818	† SGS		
		LM140-15	† Motorola (3531)				L7818C	SGS		
		LM340-15	Motorola (3531)				SG7817AP	† SiliconG		
		MC7815	† Motorola (3531)				SG7818ACP	† SiliconG		
							SG7818AP	SiliconG		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Voltage Regulators (Cont'd)

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line
<b>Fixed, Positive</b>					<b>(Cont'd)</b>				
18	1000	SG7818P	SiliconG		24	300	STK507	Sanyo	
		SFC2818	Thomson-CSF			500	μA78M24C	Fairchild	
	1500	μA7818C	Fairchild	5		μA78M24M	† Fairchild	70	
		μA7818M	† Fairchild				TDD1624		ITT
		HA17818P	Hitachi				MC78M24C		Motorola
		LAS1518	Lambda				μPC78M24		NEC
		<b>LM140-18</b>	† Motorola (3531)				SG140-24T		† SiliconG
		<b>LM340-18</b>	Motorola (3531)				SG340-24T		SiliconG
		<b>MC7818</b>	† Motorola (3532)				SG7824ACT		SiliconG
		<b>MC7818A</b>	† Motorola (3532)				SG7824AT		† SiliconG
		<b>MC7818AC</b>	Motorola (3532)				SG7824CT		SiliconG
		<b>MC7818C</b>	Motorola (3532)				SG7824T		† SiliconG
		SG140-18K	SiliconG				μA78M24C		TI
		SG140-18R	SiliconG				μA78M24M		† TI
		SG340-18K	SiliconG		15	1000	μPC7824		NEC
		SG340-18R	SiliconG				STK523		Sanyo
		SG7818ACK	SiliconG				L7824		† SGS
		SG7818ACP	SiliconG			L7824C	SGS		
		SG7818AK	† SiliconG			SG7824ACP	SiliconG		
		SG7818CK	SiliconG	20		SG7824AP	† SiliconG		
		SG7818CP	SiliconG			SG7824CP	SiliconG		
		SG7818K	† SiliconG			SG7824P	† SiliconG		
		SG78R18	SiliconG			SFC2824	Thomson-CSF		
		μA7818C	TI		1500	μA7824C	Fairchild		
	2000	MC78T18	† Motorola	25		μA7824M	† Fairchild		
		MC78T18C	Motorola			LAS1524	Lambda		
		STK532	Sanyo			<b>LM140-24</b>	† Motorola (3531)		
		L78S18	† SGS			<b>LM340-24</b>	Motorola (3531)		
		L78S18C	SGS			<b>MC7824</b>	† Motorola (3532)		
	3000	SG153-18	† SiliconG	30		<b>MC7824A</b>	† Motorola (3532)		
		SG253-18	SiliconG			<b>MC7824AC</b>	Motorola (3532)		
		SG353-18	SiliconG			<b>MC7824C</b>	Motorola (3532)		
	8000	42050-188	Micropac			SG140-24K	SiliconG		
	12000	42055-1812	Micropac			SG140-24R	SiliconG		
		42055-1812	Toshiba	35		SG340-24K	SiliconG		
20	100	TA78L020	Toshiba			SG340-24P	SiliconG		
		TA78L020A	Toshiba			SG340-24R	SiliconG		
	500	MC78M20C	Motorola			SG7824ACK	SiliconG		
		SG140-20T	† SiliconG	40		SG7824ACR	SiliconG		
		SG340-20T	SiliconG			SG7824AK	† SiliconG		
		SG7820ACT	SiliconG			SG7824CK	SiliconG		
		SG7820AT	† SiliconG			SG7824CR	SiliconG		
		SG7820CT	SiliconG			SG7824K	† SiliconG		
		SG7820T	† SiliconG			SG78R24	SiliconG		
		μA78M20C	TI	45		μA7824C	TI		
		μA78M20M	† TI		2000	HA17824P	Hitachi		
	1000	LAS1520	Lambda			MC78T24	† Motorola		
		SG140-20K	† SiliconG			MC78T24C	Motorola		
		SG140-20R	† SiliconG			STK533	Sanyo		
		SG240-20K	SiliconG			L78S24	† SGS		
		SG340-20P	SiliconG	50		L78S24C	SGS		
		SG340-20R	SiliconG		4000	42050-244	Micropac		
		SG7820K	† SiliconG		10000	42055-2410	Micropac		
		SG7820R	† SiliconG		26	42050-264	Micropac		
		SG78R20	† SiliconG	55	8000	42055-2608	Micropac		
	2000	CJSE009	Solitron		28	LAS1528	Lambda		
		CJSE011	Solitron		4000	42050-284	Micropac		
		CJSE013	Solitron		8000	42055-2808	Micropac		
	8000	42050-208	Micropac		30	42050-304	Micropac		
		42055-2010	Micropac	60	31	Shunt	TAA550		
22	4000	42050-224	Micropac		32	4000	42050-324		
	10000	42055-2210	Micropac		33	Shunt	TAA550		
24	100	MC78L24AC	Motorola	65	34	4000	42050-344		
		MC78L24C	Motorola		35	Shunt	TAA550		
		TA78L024	Toshiba						
		TA78L024A	Toshiba						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Voltage Regulators (Cont'd)

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line						
<b>Fixed, Negative</b>					5.2	1500	(Cont'd)								
2	1000	MC7902C	Motorola				SG320-5	SiliconG	70						
	1500	LAS1802	Lambda				SG320-5	SiliconG							
	2000	$\mu$ A79E02	† Fairchild				SG7905	† SiliconG							
		$\mu$ A79E02C	Fairchild				SG7905	† SiliconG							
3	100	MC79L03AC	Motorola	5			SG7905	SiliconG							
		MC79L03C	Motorola				SG7905	SiliconG							
5	100	MC79L05AC	Motorola	10			LM145K5	† National	75						
		MC79L05C	Motorola				LM345K5	National							
		LM320L05	National				$\mu$ A79M06C	TI	80						
		LM79L05	National				$\mu$ A79M06M	† TI							
		LM79L05A	National				LAS1806	Lambda	80						
	MC79L05AC	TI	MC7906C				Motorola (3531)								
	MC79L05C	TI	$\mu$ A7906C				TI	80							
	250	LM320ML05	National				2000		CJSE018	Soliton					
	500	$\mu$ A79M05C	Fairchild				5000	CJSE020	Soliton	85					
		$\mu$ A79M05M	† Fairchild				42051-065	Micropac							
1000	1500	LM120H5	† National	7	5000	42051-075	Micropac	85							
		LM320H5	National	20	500		$\mu$ A7908M		† Fairchild						
		LM320MP5	National				$\mu$ A79M08C		Fairchild						
		LM79M05	National				SG120-08T		† SiliconG						
		SG120-05T	† SiliconG				SG320-08T		SiliconG						
		SG220-05T	SiliconG				$\mu$ A79M08C		TI						
		SG320-05T	SiliconG				$\mu$ A79M08M		† TI						
		$\mu$ A79M05C	TI				$\mu$ A7908C		Fairchild	95					
		$\mu$ A79M05M	† TI						LAS1808		Lambda				
		$\mu$ PC7905	NEC				MC7908C		Motorola (3531)	95					
		SG120-05P	† SiliconG				$\mu$ PC7908		NEC						
		SG220-05P	SiliconG				SG120-08K		† SiliconG	100					
		SG320-05P	SiliconG				SG120-08R		† SiliconG						
		1500	1500				$\mu$ A7905C		Fairchild	30	5000		SG320-08K	SiliconG	100
							$\mu$ A7905M		† Fairchild				SG320-08P	SiliconG	
LAS1805	Lambda						SG320-08R	SiliconG	100						
LAS18A05	Lambda			SG7908	† SiliconG										
MC7905C	Motorola (3531)			SG7908A	† SiliconG	105									
LM120K5	† National			SG7908AC	SiliconG										
LM320K5	National			SG7908C	SiliconG	105									
LM320T5	National			$\mu$ A7908C	TI										
LM7905	National			42051-085	Micropac	105									
SG120-05K	† SiliconG			42051-095	Micropac										
SG220-05K	SiliconG			LAS1810	Lambda	110									
SG220-05R	SiliconG			42051-105	Micropac										
SG320-05K	SiliconG			12	100	MC79L12AC	Motorola	110							
SG320-05R	SiliconG						MC79L12C		Motorola						
SG7905	† SiliconG			LM320L12	National	115									
SG7905A	† SiliconG	LM79L12	National												
SG7905C	SiliconG	LM79L12A	National	115											
$\mu$ A7905C	TI	MC79L12AC	TI												
TDC2905	Thomson-CSF	MC79L12C	TI	115											
UC7905	† Unitrade (3711)	LM120H12	† National												
UC7905A	† Unitrade (3711)	LM320H12	National	120											
UC7905AC	Unitrade (3711)	SG120-12T	† SiliconG												
UC7905C	Unitrade (3711)	SG220-12T	SiliconG	120											
		SG320-12T	SiliconG												
3000	5000	MC79052C	Motorola	55	500		$\mu$ A79M12C	Fairchild	125						
		LM145K5	† National				$\mu$ A79M12M	† Fairchild							
		LM345K5	National				LM320MP12	National							
5.2	500	LM79M12C	National	60			LM79M12	National	125						
		LM79L12C	Motorola				$\mu$ A79M12C	TI							
		LM79L12A	National				$\mu$ A79M12M	† TI							
		MC79L12AC	TI				LM120K12	† National							
1000	1500	MC7905	Motorola (3531)	65	1000		LM320K12	National	130						
		SG120-5	SiliconG				LM320T12	National							
		SG220-5	SiliconG				$\mu$ PC7912	NEC							
		SG320-5	SiliconG				SG120-12P	† SiliconG							
		SG220-5	SiliconG				SG220-12P	SiliconG							
SG220-5	SiliconG	SG320-12P	SiliconG												

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Voltage Regulators (Cont'd)

Output Voltage, V	Output Current	Device	Source	Line	Output Voltage, V	Output Current	Device	Source	Line	
<b>Fixed, Negative</b>					<b>(Cont'd)</b>					
12	1500	$\mu$ A7912C	Fairchild	5	15	1500	$\mu$ A7915C	TI	70	
		$\mu$ A7912M	† Fairchild					TDC2915		Thomson-CSF
		LAS1812	Lambda					<b>UC7915</b>		† <b>Unitrade</b> (3711)
		LAS18A12	Lambda					<b>UC7915A</b>		† <b>Unitrade</b> (3711)
		<b>MC7912AC</b>	<b>Motorola</b> (3531)					<b>UC7915AC</b>		<b>Unitrade</b> (3711)
		<b>MC7912C</b>	<b>Motorola</b> (3531)					<b>UC7915C</b>		<b>Unitrade</b> (3711)
		LM7912	National				2000	42051-154		Micropac
		SG120-12K	† SiliconG					CJSE002		Solitron
		SG220-12K	SiliconG					CJSE004		Solitron
		SG220-12R	SiliconG					CJSE006		Solitron
		SG320-12K	SiliconG		10	16	4000	42051-164		Micropac
		SG320-12R	SiliconG			18	100	MC79L18C		Motorola
		SG7912	† SiliconG				500	SG120-18T		† SiliconG
		SG7912A	† SiliconG					SG320-18T		SiliconG
		SG7912AC	SiliconG			15	1000	$\mu$ PC7918		NEC
		SG7912C	SiliconG			1500	LAS1818	Lambda		
		$\mu$ A7912C	TI				<b>MC7918C</b>	<b>Motorola</b> (3532)		
		TDC2912	Thomson-CSF				MC79L18AC	Motorola		
		<b>UC7912</b>	† <b>Unitrade</b> (3711)				SG120-18K	† SiliconG		
		<b>UC7912A</b>	† <b>Unitrade</b> (3711)	20			SG120-18R	† SiliconG		
		<b>UC7912AC</b>	<b>Unitrade</b> (3711)				SG320-18K	SiliconG		
		<b>UC7912C</b>	<b>Unitrade</b> (3711)				SG320-18P	SiliconG		
	4000	42051-124	Micropac				SG320-18R	SiliconG		
14	4000	42051-144	Micropac				SG7918	† SiliconG		
15	100	MC79L15AC	Motorola	25			SG7918A	† SiliconG		
		MC79L15C	Motorola				SG7918AC	SiliconG		
		LM320L15	National				SG7918C	SiliconG		
		LM79L15	National				$\mu$ A7918C	TI		
		LM79L15A	National			4000	42051-184	Micropac		
		MC79L15AC	TI		30					
		MC79L15C	TI		20	500	SG120-20T	† SiliconG		
	200	LM120H15	† National				$\mu$ A79M20C	TI		
		LM320H15	National				$\mu$ A79M20M	† TI		
		SG120-15T	† SiliconG			1500	LAS1820	Lambda		
		SG220-15T	SiliconG				SG120-20K	† SiliconG		
		SG320-15T	SiliconG	35			SG120-20R	† SiliconG		
	250	LM320ML15	National				SG7920	† SiliconG		
	500	$\mu$ A79M15C	Fairchild				SG7920A	† SiliconG		
		$\mu$ A79M15M	† Fairchild				SG7920AC	SiliconG		
		LM320MP15	National			2000	SG7920C	SiliconG		
		LM79M15	National				CJSE010	Solitron		
		$\mu$ A79M15C	TI				CJSE012	Solitron		
		$\mu$ A79M15M	† TI				CJSE014	Solitron		
	1000	MC79L15A	Motorola			3000	42051-204	Micropac		
		LM120K15	† National	45	22	3000	42051-223	Micropac		
		LM320K15	National				MC79L24AC	Motorola		
		LM320T15	National				MC79L24C	Motorola		
		$\mu$ PC7915	NEC				$\mu$ A79M24C	TI		
		SG120-15R	† SiliconG				$\mu$ A79M24M	† TI		
		SG220-15R	SiliconG		50		$\mu$ PC7924	NEC		
		SG320-15R	SiliconG			1000	LAS1824	Lambda		
	1500	$\mu$ A7915C	Fairchild			1500	<b>MC7924C</b>	<b>Motorola</b> (3532)		
		$\mu$ A7915M	† Fairchild				$\mu$ A7924C	TI		
		LAS1815	Lambda				42051-243	Micropac		
		LAS18A15	Lambda	55			42051-263	Micropac		
		<b>MC7915AC</b>	<b>Motorola</b> (3531)				LAS1828	Lambda		
		<b>MC7915C</b>	<b>Motorola</b> (3531)				42051-283	Micropac		
		LM7915	National				42051-303	Micropac		
		SG120-15K	† SiliconG				TAA550	National		
		SG220-15K	SiliconG				42051-323	Micropac		
		SG220-15P	SiliconG				TAA550	National		
		SG320-15K	SiliconG				TAA550	National		
		SG320-15P	SiliconG				42051-363	Micropac		
		SG7915	† SiliconG							
		SG7915A	† SiliconG	65	26	3000	42051-263	Micropac		
		SG7915AC	SiliconG				LAS1828	Lambda		
		SG7915C	SiliconG				42051-283	Micropac		
							42051-303	Micropac		
							TAA550	National		
							42051-323	Micropac		
							TAA550	National		
							TAA550	National		
							42051-363	Micropac		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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LINEAR-Voltage Regulators (Cont'd)

Fixed, Dual				Adjustable, Positive						
Output Voltage, V	Output Current	Device	Source	Line	Output Voltage HI	Output Voltage Low	Output Current	Device	Source	Line
+5, +5	750, 10 1000, 50	<b>LM2935</b> LT1005C LT1005M	<b>National (3540)</b> LinearTech LinearTech		32	1.2	100 3000	TL317 LM150 LM150A LM350 LM350A	TI † LinearTech † LinearTech LinearTech LinearTech	60
±6	5000	42051	Micropac					<b>LM150</b> <b>LM250</b> <b>LM350</b>	† Motorola <b>Motorola (3532)</b> <b>Motorola (3532)</b>	65
±12	100	LM126 LM326	† National National	5				LM150 LM250 LM350 SG150 SG150A SG250 SG250A SG350 SG350A	† National National National † SiliconG SiliconG SiliconG SiliconG SiliconG	70
±15	100	XR1468 XR1568 XR4195 MC1468 MC1568 LM125 LM325 RC4195 RM4195 SG1468 SG1568 SG4501 TA7179	Exar † Exar Exar Motorola † Motorola † National National Raytheon † Raytheon SiliconG † SiliconG SiliconG Toshiba	10 15				LM350 <b>UC150</b> <b>UC250</b> <b>UC350</b>	TI † Uniredo <b>Uniredo (3711)</b> <b>Uniredo (3711)</b>	75
	200	SG1501A SG2501A SG3501A	† SiliconG SiliconG SiliconG	20			5000	LM138 LM338 LT138A LT338A LM138 LM238 LM338	† LinearTech LinearTech † LinearTech LinearTech † National National National	80
						2.5	200 500	<b>MC1469</b> MC1469	<b>Motorola (3532)</b> Motorola	85
						3.3	200	TA7089	Toshiba	
					33	1.2	3000 5000	LLM350 LLM338	Lambda Lambda	
					36	1.7 2.5	100 100	<b>CA3085A</b> μA431 TL431C TL431M TL431C TL431M	† RCA Fairchild Motorola † Motorola TI † TI	<b>(3602)</b>  <b>(3535)</b> <b>(3535)</b>
						2.85	2000	TDA200	Thomson-CSF	
16	1.3 4	40 8000	<b>ICL7663</b> LAS39U	<b>Intersil (3505)</b> Lambda	25					
20	1.2 2	10000 20	LM196 LM396 LM300 SG300 SFC2300	National National Intersil SiliconG Thomson-CSF	30					
24	3	150	LM2931	National						
26	1.8	12	<b>CA3085</b>	† RCA <b>(3602)</b>			500	SG117 SG217 SG317 TL117L	† SiliconG SiliconG SiliconG † TI	
27	0	200	LH0075 LH0075C	† National National	35					
30	2	20	LM100 SG100 SG200 SFC2100 SFC2200	† Intersil † SiliconG SiliconG † Thomson-CSF Thomson-CSF	40		1500	μA117 μA217 μA317 LLM317 LM117 LM317 LT117A LT317A MCELM117 MCELM217 MCELM317	† Fairchild Fairchild Fairchild Lambda † LinearTech LinearTech † LinearTech LinearTech MCE MCE MCE	105 110 115
	2.6	3000	LAS14AU	Lambda				<b>LM117</b> <b>LM217</b> <b>LM317</b>	† Motorola Motorola Motorola	<b>(3532)</b> <b>(3532)</b> <b>(3532)</b>
	3	Shunt 1800	TL430C L200	TI SGS				LM117 LM217 LM317 SG117 SG217 SG317	† National National National † SiliconG SiliconG SiliconG	
	4	1500 2000 5000	LAS15U LAS16U LAS19U	Lambda Lambda Lambda	45			LM217 LM317 TI	TI TI	
	4.5	12	μA305C LM305 LM305 μPC141 μPC305 μPC325 SG305	Fairchild Intersil National NEC NEC NEC SiliconG	50					
	5	500 1000 5000	μA78MG μA78G μ78HG	Fairchild Fairchild Fairchild	55					

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

(Continued)

LINEAR

Master Selection Guide

LINEAR-Voltage Regulators (Cont'd)

Adjustable, Positive						(Cont'd)					
Output Voltage Hi	Low	Output Current	Device	Source	Line	Output Voltage Hi	Low	Output Current	Device	Source	Line
37	1.2	1500	TDC0117	† Thomson-CSF		125	1.25	700	TL783C	TI	
			TDC0137	† Thomson-CSF		1000	0	100	MC1466	Motorola	
			UC117	† Unitrode (3711)					MC15666	† Motorola	
			UC217	Unitrode (3711)		<b>Adjustable, Negative</b>					
			UC317	Unitrode (3711)	5	16	1.3	40	ICL7664	Intersil (3505)	
	2	150	μA723C	Fairchild		24	2.2	5000	μA79HG	Fairchild	70
			μA723M	† Fairchild		27	0	200	LH0076	† National	
			μA723	Intersil					LH0076C	National	
			MC1723	† Motorola (3532)		30	0.035	20	LM304	National	
			MC1723C	Motorola (3532)	10				μPC142	NEC	
			LM723	† National					SG304	SiliconG	75
			LM723C	National					SFC2304	Thomson-CSF	
			RC723	Raytheon			2.2	500	μA79MG	Fairchild	
			RM723	† Raytheon				1000	μA79G	Fairchild	
			CA723	† RCA (3602)	15		2.6	1500	LAS18U	Lambda	
			CA723C	RCA (3602)		32	3.8	250	MC1463G	Motorola (3532)	80
			SA723C	Signetics				500	TA7085A	Toshiba	
			μA723	† Signetics		33	3.6	250	MC1563G	Motorola (3532)	
			μA723C	Signetics		34	3.8	600	MC1463R	Motorola (3532)	
			SG723	† SiliconG	20	37	1.2	500	LM137	† LinearTech	85
			SG723C	SiliconG					LM337	LinearTech	
			μA723C	TI					LT137A	† LinearTech	
			μA723M	† TI					LT337A	LinearTech	
			SFC2723C	† Thomson-CSF					LM137	† Motorola (3532)	
			SFC2723EC	Thomson-CSF	25				LM137M	Motorola (3532)	
	2.5	250	LM117M	† Motorola (3532)					LM237	Motorola (3532)	90
			LM217M	Motorola (3532)					LM237M	Motorola (3532)	
			LM317M	Motorola (3532)					LM337	Motorola (3532)	
			MC1569	† Motorola (3532)					LM337M	Motorola (3532)	
		600	MC1569	† Motorola	30			1500	MCELM137	MCE	95
	5	25	μA376C	Fairchild					MCELM237	MCE	
			LM376	National					MCELM337	MCE	
38	2	100	SG3532	SiliconG					LM137	† National	100
		150	LAS1000	Lambda					LM237	National	
			LAS723	Lambda	35				LM337	National	
	3	150	L123CB	SGS					LT137A	† National	
			L123CT	SGS					LT337A	National	
40	0	150	LAS3700	Lambda					SG137	† SiliconG	105
	4.5	12	μA105M	† Fairchild					SG237	SiliconG	
			LM105	† National	40				SG337	SiliconG	
			LM205	National					LM137	† TI	
			SG105	† SiliconG					LM237	TI	
			SG205	SiliconG					LM337	TI	
			SFC2105	† Thomson-CSF					UC137	† Unitrode (3711)	110
			SFC2205	Thomson-CSF	45				UC237	Unitrode (3711)	
		45	μA305AC	Fairchild					UC337	Unitrode (3711)	
			LM305A	National					MC1536R	† Motorola	
			SG305A	SiliconG		40	0.015	20	LM104	† National	115
			SFC2305	Thomson-CSF					LM204	National	
46	1.7	100	CA3085B	† RCA (3602)	50				SG204	SiliconG	
48	2	100	SG1532	† SiliconG					SG304	SiliconG	
			SG2532	SiliconG					SFC2204	Thomson-CSF	
		150	LAS1100	Lambda					SGC2104	† Thomson-CSF	
			LAS723B	Lambda		47	1.2	500	LM137HV	† LinearTech	120
56	8	1000	CJCA001	Solitron					LM337HV	LinearTech	
			CJCA007	Solitron					LT137AHV	† LinearTech	
57	1.2	500	LM117HV	† LinearTech					LT337AHV	LinearTech	
			LM317HV	LinearTech					LM137HV	† LinearTech	125
			LT17AHV	† LinearTech					LM337HV	LinearTech	
			LT317AHV	LinearTech					LT137AHV	† LinearTech	
		1500	LM117AHV	† LinearTech	60				LT337AHV	LinearTech	
			LM317AHV	LinearTech		56	8	1000	CJCA002	Solitron	
			LT117AHV	† LinearTech					CJCA008	Solitron	
77	2	150	L146	SGS		57	1.2	100	LM337L	National	
			TDB1146	Thomson-CSF	65			1500	LM137	† National	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Voltage Regulators (Cont'd)

Adjustable, Negative					(Cont'd)						
Output Voltage	Low	Output Current	Device	Source	Line	Output Voltage	Low	Output Current	Device	Source	Line
57	1.2	1500	LM237 LM337 UC137 UC237 UC337	National National † Unitrode Unitrode Unitrode	5	Switching Regulator Circuits					(Cont'd)
<b>Adjustable, Dual</b>						MC3520 † Motorola (3533)					
± 15.5	± 14.5	100	MC1468 MC1568	Motorola † Motorola		881525A † Motorola (3533)					65
± 20	± 8	100	MC1468 MC1568 SG1468 SG1568	Motorola † Motorola Silicon6 † Silicon6	10	881526 Motorola (3533)					
	± 14.5	100	SG1468T SG1568T	Silicon6 † Silicon6		881527A † Motorola (3533)					70
± 28	± 10	100	SG1502 SG2502 SG3502 SG4501	† Silicon6 Silicon6 Silicon6 Silicon6	15	882526 Motorola (3533)					
± 42	± 0.01	100 200	XR4194M MCE4194 RC4194 RM4194 SG4194 SG4194C	† Exar MCE Raytheon † Raytheon † Silicon6 Silicon6	20	882527A Motorola (3533)					
<b>Switching Regulators</b>						TL494C Motorola (3533)					
Switching Regulator Circuits						TL494M † Motorola (3533)					
			AM6301	AMD (3315)		TL495C Motorola (3533)					
			XR1524	† Exar (3381)		TL495M † Motorola (3533)					
			XR1525A	† Exar (3381)		μA78540 Motorola					
			XR1527A	† Exar (3381)		LH1605 † National					
			XR2230	Exar		LH1605C National					
			XR2235	Exar		LM1524 † National					
			XR2524	Exar (3381)		LM2524 National					
			XR2525A	Exar (3381)		LM3524 National					
			XR2527A	Exar (3381)		RC4191 Raytheon (3588)					
			XR3524	Exar (3381)		RC4192 Raytheon (3588)					85
			XR3525A	Exar (3381)		RC4193 Raytheon (3576, 3588)					
			XR3527A	Exar (3381)		RM4191 † Raytheon					
			XR494	Exar (3380)		RM4192 † Raytheon					
			XR495	Exar (3380)		RM4193 Raytheon					
			SH1605A	Fairchild		CA1524 † RCA (3602)					90
			μA494	Fairchild		CA2524 RCA (3602)					
			μA494M	† Fairchild		CA3524 RCA (3602)					
			μA78540	Fairchild		TDA4600-2 Siemens					
			μA78540M	† Fairchild		TDA4700 Siemens					
			ZN1060E	† Ferranti		TDA4700A Siemens					
			ZN1066E	Ferranti		TDA4718 Siemens					
			ZN1066J	† Ferranti		TDA4718A Siemens					
			MB3759	Fujitsu		NE5560 Signetics (3652)					
			MB3760	Fujitsu		NE5561 Signetics					
			LAS3800	Lambda		SE5560 † Signetics (526, 3652)					100
			LAS3820	Lambda		SE5561 † Signetics					
			LAS3840	Lambda		883524 Signetics (3661)					
			LAS6300	Lambda		SG1524 † Silicon6					
			LAS6300L	Lambda		SG1525A † Silicon6					
			LAS6301	Lambda		SG1526 † Silicon6					105
			LAS6302	Lambda		SG1527A † Silicon6					
			LAS6330	Lambda		SG2524 Silicon6					
			LAS6331	Lambda		SG2525A Silicon6					
			LAS6332	Lambda		SG2526 Silicon6					
			MC33063	Motorola (3533)		SG2527A Silicon6					
			MC34060	Motorola (3533)		SG3524 Silicon6					
			MC34063	Motorola (3533)		SG3525A Silicon6					
			MC3420	Motorola (3533)		SG3526 Silicon6					
			MC35060	Motorola (3533)		SG3527 Silicon6					
			MC35063	† Motorola (3533)		PWM125A † Siliconix (3081)					
				(Continued)		PWM125C Siliconix (3081)					
						PWM127 Siliconix (3081)					
						PWM127A † Siliconix (3081)					
						PWM127C Siliconix (3081)					
						PWM25A † Siliconix (3081)					120
						PWM25C Siliconix (3081)					
						PWM27A † Siliconix (3081)					
						PWM27C Siliconix (3081)					
						ULN-8126 Sprague					
						ULN-8160 Sprague					
						ULN-8161 Sprague					
						ULQ-8126 Sprague					
						ULS-8126 † Sprague					
						ULS-8160 † Sprague					130
						MC34060 TI					
						SG1524 † TI					

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Voltage Regulators (Cont'd)

Output Voltage					Output Voltage							
Hi	Low	Output Current	Device	Source	Line	Hi	Low	Output Current	Device	Source	Line	
<b>Switching Regulators</b>					<b>(Cont'd)</b>					<b>Switching Regulator Circuits</b>		<b>(Cont'd)</b>
Switching Regulator Circuits					(Cont'd)							
			SG2524	TI					UC3840	Unitrode		
			SG3524	TI					UC3842	Unitrode	70	
			SG3525A	TI					<b>UC493A</b>	† Unitrode	(3709)	
			SG3527A	TI					<b>UC493AC</b>	Unitrode	(3709)	
			TL493C	TI	5				<b>UC494A</b>	† Unitrode	(3709)	
			TL494C	TI					<b>UC494AC</b>	Unitrode	(3709)	
			TL494M	† TI					<b>UC495</b>	Unitrode	(3709)	
			TL495C	TI					<b>UC495A</b>	† Unitrode	(3709)	
			TL497AC	TI					<b>UC495AC</b>	Unitrode	(3709)	
			TL497AM	† TI	10				<b>UC495B</b>	† Unitrode	(3709)	
			<b>TL593C</b>	TI	(541)				<b>UC495BC</b>	Unitrode	(3709)	
			<b>TL594C</b>	TI	(541)							
			<b>TL594M</b>	† TI	(541)							
			TL595C	TI								
			TEA1001	Thomson-CSF	15							
			UAA4001	Thomson-CSF								
			UAA4003	Thomson-CSF								
			UAA4006	Thomson-CSF								
			PIC600	Unitrode								
			PIC601	Unitrode	20							
			PIC602	Unitrode								
			PIC610	Unitrode								
			PIC611	Unitrode								
			PIC612	Unitrode								
			PIC625	Unitrode	25							
			PIC626	Unitrode								
			PIC627	Unitrode								
			PIC635	Unitrode								
			PIC636	Unitrode								
			PIC637	Unitrode	30							
			PIC645	Unitrode								
			PIC646	Unitrode								
			PIC647	Unitrode								
			PIC655	Unitrode								
			PIC656	Unitrode	35							
			PIC657	Unitrode								
			PIC660	Unitrode								
			PIC661	Unitrode								
			PIC662	Unitrode								
			PIC670	Unitrode	40							
			PIC671	Unitrode								
			PIC672	Unitrode								
			PIC730	Unitrode								
			PIC740	Unitrode								
			PIC800	Unitrode	45							
			PIC801	Unitrode								
			PIC810	Unitrode								
			PIC811	Unitrode								
			<b>UC1524</b>	Unitrode	(3709)							
			<b>UC1524A</b>	Unitrode	(3709,3709)	50						
			<b>UC1525A</b>	Unitrode	(3709)							
			<b>UC1526</b>	Unitrode	(3709)							
			<b>UC1527A</b>	Unitrode	(3709)							
			UC1535	Unitrode								
			<b>UC1840</b>	Unitrode	(3709)	55						
			<b>UC1842</b>	† Unitrode	(3709)							
			UC2524	Unitrode								
			UC2524A	Unitrode								
			UC2525A	Unitrode								
			UC2526	Unitrode	60							
			UC2527A	Unitrode								
			UC2840	Unitrode								
			UC2842	Unitrode								
			UC3524	Unitrode								
			UC3524A	Unitrode	65							
			UC3525A	Unitrode								
			UC3526	Unitrode								
			UC3527A	Unitrode								
					(Continued)							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Other Devices

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices</b>											
AC Amplifier, Quad, Single Supply	CA3048	† RCA		Analog Shift Register (Bucket Brigade Device) Dual 5 12 stage, acts as a variable delay line in the audio frequency range)	MN3001 MN3010	Panasonic Panasonic	40	Balanced Modulator/Demodulator	(Cont'd)		85
AC Detector (Detects presence or absence of arc signals, includes adjustable threshold and time delay.)	3050	Intech		Analog Shift Register (Bucket Brigade Device) (Dual 64 stage, acts as a variable delay line in the audio range)	MN3003	Panasonic (3559)		S042	Siemens		
AC Motor Speed Control	TDA 1285	Motorola		Analog Shift Register (Bucket Brigade Device, 1024 stage)	MN3007 MN3204 MN3207	Panasonic (3559) Panasonic Panasonic		MC1496	Signetics		
AC Phase Control	TEA 1087 U111	Telefunken Telefunken	5	Analog Shift Register (Bucket Brigade Device, 128 Stage)	MN3006	Panasonic	45	MC1596	† Signetics		90
Active Filter, Resistor Programmable	R5621 R5622	Reticon Reticon		Analog Shift Register (Bucket Brigade Device, 2048 stage)	MN3008	Panasonic (3559)		TCA240	Signetics		
Active Filter, Universal	R5620	Reticon		Analog Shift Register (Bucket Brigade Device, 256 Stage)	MN3009	Panasonic (3559)		TDA0820T	Signetics		
Active Filters (See also Linear - Telecommunications Circuits)	ATF76	Barr-Brown (2852)		Analog Shift Register (Bucket Brigade Device, 328 stage)	MN3011	Panasonic (3559)		SG 1496	SiliconG		
	UAF11	Barr-Brown (2852)	10	Analog Shift Register (Bucket Brigade Device) (3,5, 190 stages)	MN3012	Panasonic (3559)	50	SG 1596	SiliconG		
	UAF11H	† Barr-Brown (2852)		Analog Shift Register (Bucket Brigade Device, 4096 stage)	MN3005	Panasonic (3559)		TA7320	Toshiba		
	UAF21	Barr-Brown (2852)		Analog Shift Register (Bucket Brigade Device, 5 12 stage)	MN3002 MN3004	Panasonic (3559) Panasonic (3559)		Blowout Resistant Transistor (simulates 40 V transistor with special protection)			
	UAF21H	† Barr-Brown (2852)		Analog Shift Register (Bucket Brigade Device, 5 12 stage)	MN3004	Panasonic (3559)		LM195	† National		95
	UAF31	Barr-Brown (2852)		Analog Shift Register Clock Generator/Driver	MN3101	Panasonic	55	LM295	National		
	UAF41	Barr-Brown (2852)	15	Analog Shift Register (Quad 64 stage)	CCD64	MicroTech		LM395	National		
	CH1290	Cermetek		Analog Shift Register (185 stage, 190x2 stage)	S10110	AMI		Comparator, Cache Address			
	CH1295	Cermetek		Analog Shift Register (256 stage)	CCD256SERP CCD256SPS	MicroTech MicroTech		TMS2150-4	TI (4222)		
	CH1296	Cermetek		Analog Signal Processor - contains two transconductance amplifiers, two current-mode switches, output voltage buffer amplifier, and precision comparator			20	TMS2150-5	TI (4222)		
	FLT-U2	Datal (2865)		GAP-01A	PMI			TMS2150-7	TI (4222)		
	FLT-U2M	Datal (2865)		GAP-01B	† PMI			TMS2150-9	TI (4222)		
	AF100	† National		GAP-01E	PMI			Comparator, Drives Lamps, Relays or Logic Loads			
	AF150	National		GAP-01F	PMI			4082/03	Barr-Brown (2852)		100
	AF151	National		Attenuator				AN829	Panasonic		
	MF10	National		Attenuator, Dual				AN829S	Panasonic		
	MF4	National (3541)	25	Attenuator, 3 1/2-Digit BCD, digitally controlled				AD7525K AD7525T MP7525B MP7525K MP7525T	AD † AD MicroPwr MicroPwr MicroPwr		
	MF5	National		Automotive Lamp Monitor				ULN-2435A ULN-2445A ULN-2455A	Sprague Sprague Sprague		
	MF6	National		Balanced Mixer				TL442M	TI		
	R5621	Reticon		Balanced Modulator/Demodulator				AD630	AD (3361)		75
	R5622	Reticon						MC1496 MC1596 LM1496 LM1596 AN610 AN612 SL640C SL641C LS025	Motorola † Motorola National † National Panasonic Panasonic Plessey Plessey SGS		
ADC Subsystem, 5 1/2 Digit, Quad Slope Integrating	AD7555	AD (2838)	30					Balanced Modulator/Demodulator (Cont'd)			
Air Flow Sensor	SG3509	SiliconG						S042	Siemens		85
Air-Core Meter Driver	LM1819	National						MC1496	Motorola		
Alarm Circuit (Single chip detector/alarm system)	CA3164E	RCA (3604)						MC1596	† Motorola		
Alarm Circuits, Temperature	3030	Intech						LM1496	National		
Alarm Circuits, Tone Alarm (Comparator with AC output)	3010	Intech	35					LM1596	† National		
Alarm Circuits, Tri-state Alert/Alarm (LED Driver)	3020	Intech						AN610	Panasonic		
Analog Memory, Triple (three analog outputs which can be stepped up or down)	S175	Siemens						AN612	Panasonic		
Analog Multiplexer, 16-Bit, Parallel-In, Serial Out	CCD16MUX	MicroTech						SL640C	Plessey		
Analog Multiplexer, 16-Bit, Serial-In, Parallel-Out	CCD16DMUX	MicroTech						SL641C	Plessey		
								LS025	SGS		
								(Continued)			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR Master Selection Guide

LINEAR-Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				Data Acquisition System, 16 Channel or 8 Channel Differential, 12-Bit A/D (Cont'd)				Fiber Optic Receiver			
Data Acquisition System front end. 16 pseudo-differential/single ended input channels.	HI5901-2	† Harris (2977)		<b>SDM858</b>	Burr-Brown	(2849)		FOR110	Burr-Brown		85
	HI5901-5	Harris (2977)						FOR120	Burr-Brown		
Data Acquisition System, 6 Channel, Sample/Hold, Pulse Width Output $\mu$ A9708	HDAS-8MC	† Datal (2861)		Data Acquisition System, 16 Channel, Sample/Hold, 12-Bit Conversion	HDAS-16MC	Datal (2861)		FOR100B	National		
	HDAS-8MM	† Datal (2861)	5		HDAS-16MM	† Datal (2861)		FOR261F-1	National		
Data Acquisition System, 8 Channel Differential, Sample/Hold, 12-Bit Conversion	HDAS-8MC	Datal (2861)		Data Acquisition System, 16 Channel 8-Bit Conversion	DAS-952R	Datal (2865)	50	FOR261F-2	National		
	HDAS-8MM	† Datal (2861)			ADC0816C	National		FOR361B	National		
Data Acquisition System, 8 Channel Differential, 16 Channel Single Ended, Sample/Hold, 12-Bit Conversion (2 device set)	AD363K	AD			ADC0817C	† National		Fiber Optic Transmitter			90
	AD363S	† AD		Data Acquisition System, 16 Element or 8 Channel Differential, Sample/Hold, 12-Bit Conversion	DAS5712	Intech		FOT110	Burr-Brown		
	AD364J	AD (2846)			DAS5714	Intech		FOT114	Burr-Brown		
	AD364K	AD (2846)	10	Data Acquisition System, 16 Element or 8 Channel Differential, Sample/Hold, 14-Bit Conversion	DAS5714	Intech		FOT114-IR	Burr-Brown		
	AD364S	† AD (2846)			DAS5716	Intech	55	FOT120	Burr-Brown		
	AD364T	† AD (2846)		Data Acquisition System, 8 Channel (expandable), Sample/Hold, 8-Bit Conversion	DAS5716	Intech		FOT180B	National		
Data Acquisition System, 8 Channel, Sample/Hold, 12-Bit A/D	MN7120	MicroNet (3042)			MN7120H	† MicroNet (3042)		Flasher (LED)			95
	MN7140	MicroNet (3042)		Data Separator/Modified Frequency Modulation Encoder	TMD2010	Telmos		LM3909	National		
	MN7140H	† MicroNet (3042)	15	DC to DC Converter: See Regulators, Switching, under Linear - Voltage Regulators.				Fluid Detector (oscillator, balance detector)			
Data Acquisition System, 8 Channel, 8-Bit Conversion	ADC0808	National			A970	† Intech		CS166	Cherry		
	ADC0809	National		DC to DC Converter (12V to 5V)	A970	† Intech		LM1830	National		
Data Acquisition System, 8 Channel, 8-Bit Conversion, Microprocessor Compatible	AD7581A	AD (2838)			A3069	Intech		ULN-2429A	Sprague		
	AD7581B	AD (2838)		DC to DC Converter ( $\pm$ 5-16V to dual outputs)	722	Burr-Brown	60	Frequency Switch, Programmable			
	AD7581C	AD (2838)	20		724	Burr-Brown		MC3344	Motorola		
	AD7581J	AD (2838)		DC to DC Converter ( $\pm$ 5-16V to four outputs)	724	Burr-Brown		Frequency to Voltage/Voltage to Frequency Converter			
	AD7581K	AD (2838)		Deglitcher (suppresses transients at outputs of D/A)	4902	TeledyneP		VFC32	Burr-Brown	(2849, 2853)	100
	AD7581L	AD (2838)			4902-83	TeledyneP		VFC3205	† Burr-Brown	(2849)	
	MP7581S	† MicroPwr		Digital Gain Set, for setting gain of op-amps	LF13006	National (3538)	65	VFC320S	† Burr-Brown	(2849)	
	MP7581A	MicroPwr (3046)	25		LF13007	National (3538)		VFC32M	† Burr-Brown	(2849, 2853)	
	MP7581B	MicroPwr (3046)		Divider (See also Multipliers/Dividers below.)	AD535J	AD		VFC42	Burr-Brown	(2849)	105
	MP7581C	MicroPwr (3046)			AD535K	AD		VFC42M	† Burr-Brown	(2849)	
	MP7581J	MicroPwr (3046)		Driver, Control Circuit for Fast Switching Transistors	UAA4002	Thomson-CSF		VFC52	Burr-Brown	(2849)	
	MP7581K	MicroPwr (3046)						VFC52M	† Burr-Brown	(2849)	
	MP7581L	MicroPwr (3046)	30	DVM Circuits-See Interface-Analog to Digital Converters, also Digital-Special			70	VFC62	Burr-Brown	(2849)	
	MP7581T	† MicroPwr (3046)		Electronic Switching Circuit (for ignitions)	CA3165	RCA (3602)		VFC62B	Burr-Brown	(2849)	110
	MP7581U	† MicroPwr (3046)		Electro-Optic Driver/Modulator	MU115	AnalogSys		VFC62C	Burr-Brown	(2849)	
	SI520	Siliconix (3083)		Fiber Optic Data Link				VFC62S	† Burr-Brown	(2849)	
	TL520	TI		3712T/R	Burr-Brown			VFC62S	† Burr-Brown	(2849)	
Data Acquisition System, 8-Channel, 12-Bit A/D Converter with Reference, Clock, Three-State Outputs, Microprocessor Interface	HS9410J	HybridSys (2981)	35	3713T/R	Burr-Brown			VFC62S	† Burr-Brown	(2849)	
	HS9410K	HybridSys (2981)		3714T	Burr-Brown			VFC62S	† Burr-Brown	(2849)	
	HS9410S/B	† HybridSys (2981)		Fiber Optic Data Link Controller	SN76333	TI		VFC62S	† Burr-Brown	(2849)	
	HS9410T	† HybridSys (2981)			SN76334	TI		VFC62S	† Burr-Brown	(2849)	
Data Acquisition System, 12 Bit	MN7150	MicroNet (3042)		Fiber Optic Detector/Preamplifier				VFC62S	† Burr-Brown	(2849)	
Data Acquisition System, 16 Channel or 8 Channel Differential, 12-Bit A/D	AD362K	AD	40	TIED460	TI			VFC62S	† Burr-Brown	(2849)	
	AD362S	† AD		TIED461	TI			VFC62S	† Burr-Brown	(2849)	
	SDM854	Burr-Brown		TIED462	TI			VFC62S	† Burr-Brown	(2849)	
	SDM856	Burr-Brown	(2849)	TIED463	TI		80	VFC62S	† Burr-Brown	(2849)	
	SDM856J	Burr-Brown	(2849)	Fiber Optic Emitter				VFC62S	† Burr-Brown	(2849)	
	SDM856K	Burr-Brown	(2849)	FOE380B-1	National			VFC62S	† Burr-Brown	(2849)	
	SDM857	Burr-Brown	(2849)	FOE380B-2	National			VFC62S	† Burr-Brown	(2849)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR—Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				Image Sensor, Linear Self Scanning (Cont'd)				Level Detector, Precision, (with internal reference, schmitt trigger action) (Cont'd)			
Frequency to Voltage/Voltage to Frequency Converter (Cont'd)				MEL512KV Panasonic				CS450 Cherry			
9401 TeledyneS				MEL64A Panasonic				Level Detector, Precision Dual			
9402 TeledyneS				MEL864A Panasonic				CS122 Cherry			
Frequency/Tone Operated Devices—See Linear-Telecommunication Circuits				MN512 Panasonic				CS129 Cherry			
Frequency-to-Voltage Converter (precision)				RA-128x128 Reticon				CS130 Cherry			
CS2907 Cherry				RA-256x256 Reticon				Level Detector, 12-Point (for fluorescent displays)			
CS2917 Cherry				RA-32X32A Reticon				XR2276 Exar			
LM2907 National				RA-50X50A Reticon				LC7555 Sanyo			
LM2917 National				RL1024C Reticon				LC7556 Sanyo			
L290 SGS				RL1024G Reticon				Level Meter, Analog Input Drives a Bar of LED's			
L291 SGS				RL1024H Reticon				LB1405 Sanyo			
L292 SGS				RL1024S Reticon				UAA170 Siemens			
L293 SGS				RL1024SF Reticon				UAA180 Siemens			
4736 TeledyneP				RL128EC Reticon				Light Activated Switch			
Function Generator, D/A Controller				RL128G Reticon				ZNP100 Ferranti			
CY360 Cybernetic				RL128S Reticon				ZNP102 Ferranti			
Hall Effect Devices (sense magnetic field) See also Digital-Other Digital Devices				RL128SF Reticon				ZNP103 Ferranti			
DN6835 Panasonic (3558)				RL1728H Reticon				ULN-3330Y Sprague			
DN6837 Panasonic (3558)				RL2048H Reticon				Light Detector (with buffer amplifier)			
DN835 Panasonic (3558)				RL256C Reticon				U123 Telefunken			
UGN-3013T Sprague				RL256EC Reticon				Light Sensor ULN-3310 Sprague			
UGN-3019 Sprague				RL256G Reticon				Linear Filter (for sensor and control systems)			
UGN-3030 Sprague				RL32WD Reticon				MF409 AnalogSys			
UGN-3040T Sprague				RL4096 Reticon				Linear Isolator (for sensor and control systems)			
UGN-3220S Sprague				RL512C Reticon				MI900 AnalogSys			
UGN-3501 Sprague				RL512EC Reticon				Linear Regulator, High Efficiency			
UGN-3604M Sprague				RL512G Reticon				UC1834 † Unitrode (3710)			
UGN-3605M Sprague				RL512S Reticon				UC2834 Unitrode (3710)			
UGS-3019 † Sprague				RL512SF Reticon				UC3834 Unitrode (3710)			
UGS-3030 † Sprague				RL64A Reticon				Low Battery Indicator, Triggers on 3 V (for use with 3 NiCd cells)			
TL172C TI				RL64W Reticon				ICM7201 Intersil			
TL173C TI				Image Sensor, 128x1, CCD				LVDT Signal Conditioner			
Image Area Sensor, Linear Self-Scanning				TC102 TI				NE5520 Signetics			
RA100X100 Reticon				Image Sensor, 256x1, CCD				Motor, AC Motor Control Circuit			
RA128X128A Reticon				CCD111 Fairchild				HEF4752V Signetics (865,899)			
RA14X41 Reticon				CCD112 Fairchild				Motor and Solenoid Driver			
RA256X256 Reticon				Image Sensor, 488x380 Area Element, CCD				CA3169 RCA (3602)			
RL1024SFX Reticon				CCD222 Fairchild				CA3169M † RCA (3602)			
RL128SFX Reticon				Image Sensor, 1024x1, CCD				Motor Driver, Full Bridge			
RL512SFX Reticon				CCD133 Fairchild				UDN-2952 Sprague (3089)			
Image Rotator 6125A OEI				CCD134 Fairchild				PIC900 Unitrode			
Image Sensor, Area, 525-Line US TV Compatible				Image Sensor, 1728x1, CCD				Motor Driver, Half Bridge			
TC201 TI				CCD122 Fairchild				SG1635 SiliconG			
Image Sensor, Area, 625-Line European TV Compatible				CCD123 Fairchild				SG3635 SiliconG			
TC202 TI				Image Sensor, 1728x1, CCD				UDN-2935Z Sprague			
Image Sensor, CC Photodiode				TC101 TI				UDN-2950Z Sprague			
CCPD128x2 Reticon				Image Sensor, 2048x1, CCD				Motor Driver, Half H			
CCPD128x4 Reticon				CCD142 Fairchild				SN75603 TI			
CCPD128x8 Reticon				CCD143 Fairchild				SN75604 TI			
Image Sensor, Circular, Self Scanning				CCD145 Fairchild				Motor, Induction Motor Energy Saver (operates off 120 Vac line)			
R064 Reticon				TC103 TI				HV1000 Harris (3471)			
R0720B Reticon				Image Sensor, 2048x1, CCD				Motor, Induction Motor Energy Saver (operates off 240 Vac line)			
Image Sensor, Linear Self Scanning				μPD792 NEC				HV1000A Harris (3471.3471)			
CCD111 Fairchild				μPD794 NEC				Motor Speed Controller			
CCD112 Fairchild				MN8027 Panasonic				MC213 AnalogSys			
CCD122 Fairchild				Image Sensor, 3456x1, CCD				Motor Speed Regulator			
CCD123 Fairchild				CCD151 Fairchild				LS7263 LSIComp			
CCD123 Fairchild				TC104 TI				TDA1085A Motorola			
CCD133 Fairchild				Keyboard Interface, Capacitive				CA3228 RCA (3602)			
CCD134 Fairchild				KB3710 GI				Motor Speed Regulator (for small dc motors). See also Linear—Consumer Circuits.			
CCD142 Fairchild				Level Detector, Logarithmic, 10 Step Analog				MC212 AnalogSys			
CCD143 Fairchild				NSL4944 National				CS140 Cherry			
CCD145 Fairchild				Level Detector, Precision, (with internal reference, schmitt trigger action)				CS175 Cherry			
CCD151 Fairchild				CS127 Cherry				CS2907 Cherry			
MEL1024K Panasonic				CS401 Cherry							
MEL1024KV Panasonic				CS402 Cherry							
MEL128 Panasonic											

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LINEAR Master Selection Guide

LINEAR—Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				<b>Multipliers/Dividers (Cont'd)</b>				<b>Overvoltage Protector (Cont'd)</b>			
Motor Speed Regulator (for small dc motors). See also Linear—Consumer Circuits. (Cont'd)				AD533S † AD (3363)				SG1542 † SiliconG			
CS2917 Cherry				AD534J AD (3363)				SG2542 SiliconG			
μA7392 Fairchild				AD534K AD (3363)				SG3423 SiliconG			
LS7263 LSIComp (803)				AD534L AD (3363)				SG3423A SiliconG			
LM1014 National				AD534S † AD (3363)				SG3523 † SiliconG			
SL440 Plessey				AD534T † AD (3363)				SG3523A † SiliconG			
TCA900 SGS				4203 Burr-Brown (2852)				SG3542 SiliconG			
TCA910 SGS				4203S † Burr-Brown (2852)				MC3423 TI			
TCA955 Siemens				4204 Burr-Brown (2852)				Peak Detector (senses and holds peak)			
NE5520 Signetics (3635)				4204S † Burr-Brown (2852)				4085 Burr-Brown (2852)			
TDA1506 Signetics				4205 Burr-Brown (2852)				4085M † Burr-Brown (2852)			
SG1731 † SiliconG				4205S † Burr-Brown (2852)				5006 OEI			
SG2731 SiliconG				4206 Burr-Brown (2852)				5030A OEI			
SG3731 SiliconG				4213 Burr-Brown (2852,2853)				5032A OEI			
EF4443 Thomson-CSF				4213S † Burr-Brown (2852,2853)				5902 † OEI			
Motor, Stepping Motor Driver				4213U Burr-Brown (2852,2853)				PKD-01 PMI			
TDA3717 MicroSci				4213V Burr-Brown (2852,2853)				PIN Diode Driver			
TDA3717 Plessey				4213W Burr-Brown (2852,2853)				DH0035 † National			
SAA1027 Signetics				4214 Burr-Brown (2852)				DH0035C National			
UDN-2949Z Sprague				4214M † Burr-Brown (2852)				UDS-5790H † Sprague			
TL376C TI				MPY100 Burr-Brown				UDS-5791H † Sprague			
Motor, Stepping Motor Translator/Driver				MPY100A Burr-Brown				TDA1074A Signetics			
UCN-4202A Sprague				MPY100S † Burr-Brown				Potentiometer, Dual Double			
UCN-4203A Sprague				RC4200 Raytheon				TDA1074A Signetics			
Multifunction Converter (generates output=(V1 ÷ V2))				RC4200A Raytheon				TDA1074A Signetics			
LH0094 National				Noise Generator (pseudo-random sequence generator for audio)				Potentiometer, 3 1/2-Digit BCD			
LH0094C National				S2688 AMI				AD7525K AD (2827,2829)			
Multifunction Convertors (XY/Z)expM				MM5837 National				AD7525L AD (2827,2829)			
4301 Burr-Brown (2852)				Oscillator, Crystal Clock (250 kHz to 60 MHz)				AD7525T † AD (2827,2829)			
4302 Burr-Brown (2852)				CK1100A Solarise				AD7525U † AD (2827,2829)			
Multipliers				CK1114A Solarise				Power Control Subsystem			
AD539J AD (3362,3363)				CK1144A Solarise				AM6300 AMD (3312)			
AD539K AD (3362,3363)				CK1145A Solarise				Power Control Zero Voltage Switch			
AD539S † AD (3362,3363)				Oscillator, Fixed Frequency (10 Hz to 20 kHz)				SL445A Plessey			
MM109 AnalogSys				4023 Burr-Brown (2852)				Power Supply Control (voltage reference, over and under voltage sensing)			
XR2208 Exar				4025 Burr-Brown (2852)				XR1543 Exar (3382)			
XR2208M † Exar				Oscillator, Quadrature				XR2543 Exar (3382)			
XR2228 Exar				4423 Burr-Brown (2852)				XR3543 Exar (3382)			
XR2228M † Exar				Overvoltage Protector				SG1543 † SiliconG			
A8495 Intech				L20V12 Lambda				SG1544 † SiliconG			
ICL8013C Intersil				L20V15 Lambda				SG2543 SiliconG			
ICL8013M † Intersil				L20V18 Lambda				SG2544 SiliconG			
MC1494 Motorola				L20V20 Lambda				SG3543 SiliconG			
MC1495 Motorola				L20V24 Lambda				SG3544 SiliconG			
MC1594 † Motorola				L20V5 Lambda				UC1543 † Unitrode (3710)			
MC1595 † Motorola				L20V6 Lambda				UC1544 † Unitrode (3710)			
CA3091 † RCA (3604)				L20V9 Lambda				UC1901 † Unitrode (3710)			
SG1402 † SiliconG				L60V12 Lambda				UC1903 † Unitrode (3710)			
SG1495 SiliconG				L60V15 Lambda				UC2543 Unitrode (3710)			
SG1595 † SiliconG				L60V24 Lambda				UC2544 Unitrode (3710)			
SG2402 SiliconG				L60V28 Lambda				UC2901 Unitrode (3710)			
SG3402 SiliconG				L60V5 Lambda				UC2903 Unitrode (3710)			
TA7158 Toshiba				L60V6 Lambda				UC3543 Unitrode (3710)			
Multipliers/Dividers				L60V28 Lambda				UC3544 Unitrode (3710)			
AD530J AD				L60V5 Lambda				UC3901 Unitrode (3710)			
AD530K AD				L60V6 Lambda				UC3903 Unitrode (3710)			
AD530L AD				MC3423 Motorola				Power Supply Controller (voltage reference, pulse generator and timing circuitry, error amp)			
AD530S † AD				MC3523 † Motorola				LA5700 Sanyo			
AD531J AD (3363)				Multipliers/Dividers (Continued)				TL496C TI			
AD531K AD (3363)				L20V12 Lambda				Power Supply Overvoltage Sensing Circuit, Pin Programmable			
AD531L AD (3363)				L20V15 Lambda				MC34062 Motorola			
AD531S † AD (3363)				L20V18 Lambda				MC35062 Motorola			
AD532J AD (3363)				L20V20 Lambda				Power Supply Overvoltage Sensing Circuit, Three-Terminal			
AD532K AD (3363)				L20V24 Lambda				MC34061 Motorola			
AD532S † AD (3363)				L20V5 Lambda				MC34061A Motorola			
AD533J AD (3363)				L20V6 Lambda				MC35061 Motorola			
AD533K AD (3363)				L20V9 Lambda				MC35061A Motorola			
AD533L AD (3363)				L60V12 Lambda				MC35061A Motorola			

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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LINEAR-Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				Reference Voltage (two terminal active circuit 1.220 V ± 5%) (Cont'd)				Reference Voltage, 2.5 V (Cont'd)			
Power Supply Supervisory Circuit	MC3324	Motorola	5	9491AM	† TeledyneS	55	LM185-2	† National	120		
	MC3324A	Motorola		9491B	TeledyneS		LM236-2	National			
	MC3424	Motorola		9491BM	TeledyneS		LM236A-2	National			
	MC3424A	Motorola					LM285-2	National			
	MC3524	Motorola					LM336-2	National			
	MC3524A	Motorola			LM385-2	National					
Power Supply Supervisory/Over-Under Voltage Protection Circuit	MC3425	Motorola	10	LM385-1	† LinearTech	60	SG1503	SiliconG	125		
	MC3425A	Motorola		LT1004C	LinearTech		SG2503	SiliconG			
	MC3525	Motorola		LT1004M	† LinearTech		SG3503	SiliconG			
	MC3525A	Motorola		LM285-1	Motorola (3535)		TDB0136	Thomson-CSF			
				LM385-1	Motorola (3535)		TDC0136	† Thomson-CSF			
Pressure Transducers, Absolute	LX04XXA	National	15	LM185	† National	65	Reference Voltage, 2.5 V (trimmable output) ZNFREF025 Ferranti				
	LX05XXA	National		Reference Voltage (two terminal active circuit) 1.8 to 5.6 V LM103 † National							
	LX05XXA0	National		Reference Voltage, 1.23 V							
	LX06XXA	National		AD589J	AD (2842)		70	Reference Voltage, 2.5, 5.0, 7.5, and 10.0 V			
	LX14XXA	National		AD589K	AD (2842)			AD584J	AD (2842)		
	LX14XXAF	National		AD589L	AD (2842)			AD584K	AD (2842)		
	LX14XXAS	National		AD589M	AD (2842)			AD584L	AD (2842)		
	LX1600A	National		AD589S	† AD (2842)			AD584S	† AD (2842)		
	LX16XXA	National		AD589T	† AD (2842)			AD584T	† AD (2842)		
	LX18XXA	National		AD589U	† AD (2842)			AD584U	† AD (2842)		
Pressure Transducers, Backward Gage	LX0603GB	National	20	ICL8069	Intersil (3521)	75		Reference Voltage, 4 V (trimmable output) ZNFREF040 Ferranti			
	LX06XXGB	National		ICL8069M	† Intersil (3521)			Reference Voltage, 5 V			
	LX16XXGB	National		LM185-1	† National			AD584	AD		
	LX18XXGB	National		LM285-1	National		MP5531	† MicroPwr (3529)			
				LM385-1	National		MP5541	† MicroPwr			
Pressure Transducers, Differential	LX16XXD	National	25	Reference Voltage, 1.26 V (two terminal active circuit) ZN423 Ferranti				80	MP5543	† MicroPwr	140
	LX18XXD	National		Reference Voltage, 2.45 V (two terminal active circuit)							
	LX96XXD	National		VR-182A	Datel	MPREF-02	† MicroPwr (3529)				
Pressure Transducers, Gage	LX06002D	National	30	VR-182B	Datel	MPREF-02A	† MicroPwr	145	MPREF-02C	MicroPwr	
	LX06002G	National		VR-182C	Datel	MPREF-02D	MicroPwr				
	LX06XXD	National		ZN404	Ferranti	MPREF-02F	MicroPwr				
	LX16XXG	National		ZN458	Ferranti	MPREF-02H	MicroPwr				
	LX18XXG	National		ZN458A	Ferranti	MC1400-5	Motorola (3534)				
Programmable Quad Comparator, Micropower	LP165	National	35	ZN458B	Ferranti	MC1400A-5	Motorola (3534)	85	MC1404-5	Motorola (3535)	
	LP365	National		Reference Voltage, 2.5 V					MC1500-5	† Motorola (3534)	
Proximity Detector, Electromagnetic	CS191	Cherry		AD1403	AD (2842)	MC1500A-5	Motorola (3534)				
	CS209	Cherry		AD1403A	AD (2842)	MC1504-5	† Motorola (3535)				
Pulse-Width Modulated Controller (for motors, heaters, lamps)	MC343	AnalogSys		AD580J	AD (2842)	LM136-5	† National				
Pulse-Width Modulator Control System (for switching regulators, motor-speed controllers)	XR2230	Exar (3382)		AD580K	AD (2842)	LM136A-5	† National				
Quantizer, 4-Bit (Quantizes analog voltage into 15 equally-spaced levels and outputs 4-Bit binary digital word at sampling rates up to 100 MHz.)	AM6688	AMD (3309)		AD580L	AD (2842)	LM168-5	† National				
	AM6688L-6	AMD (3309)		AD580S	† AD (2842)	LM236-5	National				
	AM6688L-7	AMD (3309)		AD580T	† AD (2842)	LM236A-5	National				
	AM6688L-8	AMD (3309)		AD580U	† AD (2842)	LM268-5	National				
	AM6688M-6	† AMD (3309)	AD584	AD (2842)	LM336-5	National					
	AM6688M-7	† AMD (3309)	LM136-2	† LinearTech	LM336B-5	National					
	AM6688M-8	† AMD (3309)	LM336-2	LinearTech	LM368-5	National					
Read Chain Data Comparator	TL712	TI	LM385-2	LinearTech	REF-02	† PMI	160	REF-02A	† PMI	165	
	TL721	TI	LT1004C-2	LinearTech	REF-02C	PMI					
Reference Voltage (two terminal active circuit 1.220 V ± 5%)	MP5010	† MicroPwr (3529)	LT1004M-2	† LinearTech	REF-02D	PMI					
	MP5010A	† MicroPwr (3529)	LT1009C-2	LinearTech	REF-02E	PMI					
	LM113	† National	LT1009M-2	† LinearTech	REF-02H	PMI					
	LM113-1	† National	MP5540	MicroPwr	REF-05A	PMI					
	LM113-2	National	LM285-2	Motorola (3535)	REF-05B	† PMI					
	LM313	National	LM385-2	Motorola (3535)	REF-02	Raytheon (3585)		170	REF-02A		Raytheon (3585)
			LM385B-2	Motorola (3535)	REF-02C	Raytheon (3585)					
			MC1400-2	Motorola (3534)	REF-02D	Raytheon (3585)					
			MC1400A-2	Motorola (3534)	REF-02E	Raytheon (3585)					
			MC1403	Motorola (3535)	REF-02H	Raytheon (3585)					
			MC1403A	Motorola (3535)	REF-02N	Raytheon (3585)					
			MC1500-2	† Motorola (3534)	9495A	† TeledyneS					
			MC1500A-2	† Motorola (3534)	9495C	TeledyneS					
			MC1503	† Motorola (3535)	TSC9495	TeledyneS					
			MC1503A	† Motorola (3535)	Reference Voltage, 5 V (trimmable output) ZNFREF050 Ferranti						
			LM136-2	† National	Reference Voltage, 6.1 V (trimmable output) ZNFREF061 Ferranti						
			LM136A-2	† National			180				

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR

Master Selection Guide

LINEAR-Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				Reference Voltage, 10 V (or -10 V) (Cont'd)				Resolver and Synchro Systems (Cont'd)			
Reference Voltage, 6.2 V				60				HSDC-14 DDC			
LM168-6 † National				HC2701 HyComp				HSDC-360 DDC			
LM268-6 National				HC2701-883 † HyComp				HSDC-8915 DDC			
LM368-6 National				HC2710 HyComp				HSDC-8916 DDC			
Reference Voltage, 6.25 V				65				HSDC-8917 DDC			
MC1400-6 Motorola (3534)				HC2711-883 † HyComp				HSDC-8918 DDC			
MC1400A-6 Motorola (3534)				MP5532 † MicroPwr (3529)				HSDC-8919 DDC			
MC1404-6 Motorola (3535)				MP5542 † MicroPwr				HSDC-8920 DDC			
MC1500-6 † Motorola (3534)				MP5545 † MicroPwr				HSDC-8921 DDC			
MC1500A-6 † Motorola (3534)				MPREF-01 † MicroPwr (3529)				HSDC-8922 DDC			
MC1504-6 † Motorola (3535)				MPREF-10 † MicroPwr (3529)				HXCDX-14 DDC			
Reference Voltage, 6.9 V				70				HXCT-14 DDC			
LM129 † LinearTech				MC1400-10 Motorola (3534)				HXDC-10 DDC			
LM329 LinearTech				MC1400A-10 Motorola (3534)				HXDC-14 DDC			
Reference Voltage, 6.9 V Temperature Stabilized				75				SDC-14510 DDC			
LM129 † National				MC1404-10 Motorola (3535)				SDC-14520 DDC			
LM199 † National				MC1404A-10 Motorola (3535)				SDC-19100 DDC			
LM199A † National				MC1500-10 † Motorola (3534)				Resolver, Digital to Resolver Converter (DRC)			
LM299 National				MC1500A-10 † Motorola (3535)				DRC1765 † AD			
LM299A National				MC1504-10 † Motorola (3535)				DRC1766 † AD			
LM329 National				MC1504A-10 † Motorola (3535)				80			
LM329 National				LH0070 † National				RMS to DC Converter			
LM399 National				LM168-10 † National				AD536A AD (3363)			
LM399A National				LM268-10 National				AD536AJ AD (3363)			
LM399A National				LM368-10 National				AD536AK AD (3363)			
LM399A National				REF-01 † PMI				AD536AS † AD (3363)			
LM399A National				REF-01A † PMI				AD636J AD (3359,3363)			
LM399A National				REF-01C PMI				AD637J AD (3360,3363)			
LM399A National				REF-01D PMI				AD637K AD (3360,3363)			
LM399A National				REF-01E PMI				4340 Burr-Brown			
LM399A National				REF-01H PMI				4341 Burr-Brown			
LM399A National				REF-10A PMI				(2852)			
LM399A National				REF-10B † PMI				(2852)			
LM399A National				REF-01 Raytheon (3584)				LH0091 † National			
LM399A National				REF-01A Raytheon (3584)				LH0091C National			
LM399A National				REF-01C Raytheon (3584)				Sample and Hold Circuits			
LM399A National				REF-01D Raytheon (3584)				LF198 AMD			
LM399A National				REF-01E Raytheon (3584)				LF298 AMD			
LM399A National				REF-01H Raytheon (3584)				LF398 † AMD			
LM399A National				REF-01H Raytheon (3584)				AD582K AD (2847)			
LM399A National				REF-10A Raytheon (3584)				AD582S † AD (2847)			
LM399A National				TSC9496 TeledyneS				AD583K AD (2847)			
Reference Voltage, 6.95 V				Reference Voltage, 10.24 V				AD583S † AD (2847)			
LM199 † LinearTech				LH0071 National				AD585J AD (2842,2847)			
LM399 LinearTech				LH0071-0 National				AD585S † AD (2842,2847)			
Reference Voltage, 7.5 V				Reference Voltage, ± 10 V				ADSHC-85 AD (2847)			
AD584 AD				AD2702L AD				ADSHC-85E † AD			
MP5544 MicroPwr				AD2702S † AD				ADX346 AD			
Reference Voltage, 10 V (trimmable output)				AD2702U AD				HTC-0300 AD (2847)			
REF101J Burr-Brown (2852)				AD2712K AD				HTC-0500 AD (2845,2847)			
REF101S † Burr-Brown (2852)				AD2712L AD				HTS-0025 AD (2847)			
ZNREF100 Ferranti				AD2720 AD (2842)				MH410 AnalogSys			
Reference Voltage, 10 V (or -10 V)				R675B-3 † HybridSys				MN245 Analogic			
AD2700J AD (2842)				R675C-3 HybridSys				MP250M Analogic			
AD2700L AD (2842)				HC2702 HyComp				MP260 Analogic			
AD2700S † AD (2842)				HC2702-883 † HyComp				MP261 Analogic			
AD2700U † AD (2842)				HC2712 HyComp				MP270 Analogic			
AD2701J AD				HC2712-883 † HyComp				MP271 Analogic			
AD2701L AD				Regulator Diode with Amplifier (6.8 to 11 V)				MP272 Analogic			
AD2701S † AD				MCA-series Motorola				MP282A Analogic			
AD2701U † AD				Resolver and Synchro Systems				SHC298AM Burr-Brown			
AD2710K AD (2842)				SDC1740 AD				(2849)			
AD2710L AD (2842)				SDC1741 AD				SHC80 Burr-Brown			
AD2720A AD (2842)				SDC1742 AD				(2849)			
AD2720C AD (2842)				SDC1742 AD				SHC85 Burr-Brown			
AD2720S † AD (2842)				DRC10500 DDC				(2849)			
AD2720T † AD (2842)				HMSDC-8700 DDC				SHC85ET Burr-Brown			
AD2720B AD				HRCDX-14 DDC				(2849)			
AD581J AD (2842)				HRCT-14 DDC				SHM-20C DataL (2864)			
AD581K AD (2842)				HRDC-10 DDC				SHM-20M † DataL (2864)			
AD581L AD (2842)				HRDC-14 DDC				SHM-6MM † DataL (2861)			
AD581S † AD (2842)				HSCDX-14 DDC				(Continued)			
AD581T † AD (2842)				HSCT-14 DDC							
AD581U † AD (2842)				HSDC-10 DDC							
AD584 AD											
REF101 Burr-Brown (2852)											
HSREF01 HybridSys											
R675B-1 † HybridSys											
R675B-4 † HybridSys											
R675B-5 † HybridSys											
R675C-1 HybridSys											
R675C-5 HybridSys											
HC2700 HyComp											
HC2700-883 † HyComp											

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR-Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				<b>Sample and Hold Circuits (Cont'd)</b>				<b>Switches</b>			
<b>Sample and Hold Circuits (Cont'd)</b>				CSH 101A † TeledyneC				ICL8018AC Intersil			
SHM-7MC Datel (2864)				4856 TeledyneP				ICL8018AM † Intersil 110			
SHM-7MR † Datel (2864)				4857 TeledyneP				ICL8019AC Intersil			
SHM-8MC Datel				4860 TeledyneP				ICL8019AM † Intersil			
SHM-HUMC Datel (2861)				4860-83 † TeledyneP				ICL8020AC Intersil			
SHM-HUMM † Datel				SCR/TRIAC Control (burst control)				ICL8020AM † Intersil			
SHM-IC-1 Datel (2864)				L121 SGS 75				<b>Switching Regulators. See Linear-Voltage Regulators, Switching</b> 115			
SHM-IC-1M † Datel				SCR/TRIAC Control (for push-button control)				<b>Tachometer: See Frequency to Voltage and Frequency Switch above</b>			
SHM-LM-2 Datel (2864)				UAA4007 Thomson-CSF				<b>Temperature Controlled Differential Pair</b>			
SHM-LM-2M † Datel				UAA4008 Thomson-CSF				μA726C Fairchild			
μA 198 † Fairchild				SCR/TRIAC Control (phase control)				μA726M † Fairchild			
μA298 Fairchild				L120 SGS				<b>Temperature Transducers</b>			
μAF398 Fairchild				UAA 145 Telefunken				AD590I † AD			
HA-2420 Harris (2955)				UAA 146 Telefunken				AD590J † AD			
HA-2425 † Harris (2955)				SCR/TRIAC Control, (zero voltage control)				AD590K † AD			
HA-5310 Harris				TEA 1511 Thomson-CSF				AD590L † AD			
HA-5320-2 † Harris (2958)				Sensor/Controller, Air Temperature				AD590M AD			
HA-5320-5 Harris (2958)				MS 120 AnalogSys				AD590I † Intersil			
SH-8518 DDC				Sensor/Controller, Ground Moisture and Liquid Level				AD590J † Intersil			
ASH240 Intech				MS211 AnalogSys				AD590K † Intersil			
ASH250 Intech				Sensor/Controller, Relative Humidity				AD590L † Intersil			
ASH271 Intech				MS214 AnalogSys				AD590M † Intersil			
IH5110 † Intersil				Sensor, Ultrasonic Object Detector (position and distance)				ICL8073 Intersil			
IH5111 † Intersil				MS118 AnalogSys				ICL8074 Intersil			
IH5112 Intersil				Sequencer, 10 Outputs, 100 μs to 100 Seconds				MMBTS102 Motorola			
IH5113 Intersil				MC116 AnalogSys				MMBTS103 Motorola			
IH5114 Intersil				Serial Analog Delay (analog storage units with read in/read out shift register) (See also Analog Shift Registers above)				MMBTS105 Motorola			
IH5115 Intersil				R5106 Reticon				LM134 † National			
MN343 MicroNet (3045)				Serial Analog Delay, Tapped (bucket brigade with 32 taps)				LM135 † National			
MN343H † MicroNet (3045)				TAD32A Reticon				LM234 National			
MN344 MicroNet (3045)				Servo Amplifier, for Motor Control				LM334 National			
MN344H † MicroNet (3045)				ZN409CE Ferranti				LM335 National			
MN346 MicroNet (3045)				Servo Controller, for VTR or dc Servos				LM35 National			
MN346H † MicroNet (3045)				AN6880 Panasonic				LM3911 National			
MN347 MicroNet (3045)				LB1601 Sanyo				μPC3911 NEC			
MN347H † MicroNet (3045)				Servo Controller, Proportional Control				REF-02 † PMI			
MN375A MicroNet (3045)				XR2264 Exar				REF-02A † PMI			
MN7130 MicroNet (3042)				XR2265 Exar				REF-02C PMI			
MN7130H † MicroNet (3042)				Servo Controller, Radio Controlled Cars, 2 Channel				REF-02D PMI			
LF198 † National				XR2266 Exar				REF-02E PMI			
LF198A † National				Signal Processor, Real Time Digital Processing of Analog Signals (programmable)				REF-02H PMI			
LF298 National				2920 Intel				LA7011 Sanyo			
LF298A National				2921 Intel				TDC0134 † Thomson-CSF			
LF398 National				Smoke Detectors: See Linear-Consumer Circuit, Miscellaneous				TDC0135 † Thomson-CSF			
LF398A National				Solar Transceiver				<b>Thermal Converter (matched transistors, diffused resistors)</b> 4131 Burr-Brown			
LH0023 † National				M1812 National				<b>Threshold Switch (oscillator, switch with hysteresis)</b>			
LH0023C National				Stepper Motor Controller				TCA105 Siemens			
LH0043 † National				SSI183 SiliconSys				<b>Threshold Switch (2/3 supply voltage)</b>			
LH0043C National				Stepper Motor Predriver				CS560 Cherry			
LH0053 † National				SSI184 SiliconSys				TCA345 Siemens			
LH0053C National				Stepping Motor Controller/Driver, 24 V				<b>Tone Decoder (traffic signal control and detector)</b>			
SHM6401 National				SAA1042A Motorola				LA2200 Sanyo			
μPC649 NEC				Stepping Motor Controller/Driver, 6-12 V				<b>Track and Hold</b>			
5021 OEI				SAA1042 Motorola				NTS-0010 AD (2842,2847)			
5025 OEI				Switch Driver, for Power Transistors				ADH-050 DDC			
SMP10A † PMI				SG1629 † SiliconG				ADH-051 DDC			
SMP10B † PMI				SG3629 SiliconG				TH-8530 DDC			
SMP10E PMI				Switched Capacitor Array				THA05203 DDC			
SMP10F PMI				SCA-12 SiliconSys (3685,4872)				MNO300A MicroNet (3045)			
SMP11A † PMI				SCA-6 SiliconSys (3685,4872)				MNO50 MicroNet (3045)			
SMP11B † PMI				Switched Capacitor Filter Array (mask programmable)				MN373 MicroNet (3045)			
SMP11E PMI				R5626 Reticon				MN374 MicroNet			
SMP11F PMI				70				MN376 MicroNet (3045)			
SMP11G PMI								MN377 MicroNet (3045)			
SMP11H PMI								MN379 MicroNet (3045)			
SMP11I PMI								<b>Traffic Information Control System (ARI-DK type)</b>			
SMP11J PMI								LA2211 Sanyo			
SMP11K PMI											
SMP11L PMI											
SMP11M PMI											
SMP11N PMI											
SMP11O PMI											
SMP11P PMI											
SMP11Q PMI											
SMP11R PMI											
SMP11S PMI											
SMP11T PMI											
SMP11U PMI											
SMP11V PMI											
SMP11W PMI											
SMP11X PMI											
SMP11Y PMI											
SMP11Z PMI											
LF198 † Signetics											
LF298 Signetics											
LF398 Signetics											
NE5537 Signetics (3689)											
SE5537 † Signetics											
(Continued)											

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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LINEAR-Other Devices (Cont'd)

Function	Device	Source	Line	Function	Device	Source	Line	Function	Device	Source	Line
<b>Other Linear Devices (Cont'd)</b>				<b>Voltage to Frequency/Frequency to Voltage Converter (Cont'd)</b>				<b>Dual Voltage Level Indicator: See Window Discriminator below</b>			
Transverse Low Bias FET	ter, Quad Chirped (for discrete fourier power spectral density applications)			VFC32	Burr-Brown			Quad Voltage Level Monitor/Alarm (activated if any of 4 inputs differs by more than $\pm 5$ , $\pm 10$ or $\pm 20\%$ of selected value)	ULN-2401A	Sprague	
	R5601A-1	Reticon		VFC32M	† Burr-Brown						
	R5601A-2	Reticon		VFC42	Burr-Brown						
Triac Angle	ring Control Circuit			VFC42M	† Burr-Brown						
	TDA1185	Motorola		VFC52	Burr-Brown			55	Two-Wire Bidirectional Communication System		
Undervoltage Protector/Low Voltage Monitor				VFC52M	† Burr-Brown				<b>LM1893</b>	<b>National</b>	<b>(3539)</b>
	CS188	Cherry		VFV-100K	Datel						
Video Delay, CCD				VFV-10K	Datel						
	CCD321A	Fairchild	5	<b>XR4151</b>	<b>Exar</b>	<b>(3385)</b>					
Voltage Converter, +5 V to $\pm 5$ V				XR4151M	† Exar			60			
	VI-7660C	Datel (2865)		A8400	Intech						
	VI-7660M	Datel (2865)		A8402	Intech						
	ICL7660	Intersil (3498)		A8404	Intech						
	T3C7660	TeledyneS (3097)		MCEVFC32	MCE						
Voltage Detector, Indicator, Regulator, (programmable zener)				MPVFC32B	MicroPwr			65			
	ICL8211C	Intersil (3522,3523)	10	MPVFC32K	MicroPwr						
	ICL8211M	† Intersil (3522,3523)		MPVFC32S	MicroPwr						
	ICL8212C	Intersil (3522,3523)		LM131	† National						
	ICL8212M	† Intersil (3522,3523)		LM131A	† National						
Voltage Doubler/Inverter, Converts 4.5-20V to 9 to 40V or -4.5 to -20V.	SI7661	Siliconix		LM231	National			70			
Voltage Doubler/Inverter, 1.5-10V to 3 to 20V or -1.5 to -10V.	ICL7660	Intersil (3081)	15	LM231A	National						
Voltage Level Alarm	3041	Intech		LM331	National						
Voltage Level Alarm, Quad	3040	Intech		LM331A	National						
Voltage Overvoltage Protector. (Crowbar) See Overvoltage Protector above				RC4151	Raytheon						
Voltage to Frequency Converter				RC4152	Raytheon			75			
	AD537J	AD	20	RC4153	Raytheon						
	AD537K	AD		RM4151	† Raytheon						
	AD537S	† AD		RM4152	† Raytheon						
	AD650	AD (2840)		RM4153	† Raytheon						
	ADVFG32	AD		4780	TeledyneP			80			
	VFC320	Burr-Brown (2840)	25	4781	TeledyneP						
	VFQ-1	Datel (2865)		4782	TeledyneP						
	VFQ-2	Datel (2865)		9400	TeledyneS						
	VFQ-3	Datel (2865)		9401	TeledyneS			85			
	7042H	Intech		9402	TeledyneS						
	SSM2031	SSM									
	4731	† TeledyneP		Voltage Variable Gain Block							
	4731-83	† TeledyneP		MQ328	AnalogSys						
	4732	TeledyneP		Window Discriminator (indicates voltage above, below, inside, or outside two adjustable limits)							
	4732-83	† TeledyneP	35	TCA965	Siemens						
	4733	† TeledyneP		Window Discriminator (indicates when input is above or below two limits)							
	4733-83	† TeledyneP		CS180	Cherry						
	4734	TeledyneP		Zero Voltage and Zero Crossing Triggers (triac and SCR control)							
	4734-83	† TeledyneP	40	TDA2090	MicroSci			90			
	4735	† TeledyneP		TDA2091	MicroSci						
	4735-83	† TeledyneP		CA3059	Motorola						
	4736-83	† TeledyneP		CA3079	Motorola						
	4739	TeledyneP		MC3370	Motorola						
	4739-80	† TeledyneP		UAA1016A	Motorola						
	4743	TeledyneP	45	UAA1016B	Motorola			95			
	4743-80	† TeledyneP		SL441	Plessey						
	TDB0131	Thomson-CSF		SL443	Plessey						
				SL446	Plessey						
Voltage to Frequency/Frequency to Voltage Converter				<b>CA3058</b>	† RCA (3603)			100			
	AD650J	AD (2840)		<b>CA3059</b>	† RCA (3603)						
	AD650K	AD (2840)		<b>CA3079</b>	† RCA (3603)						
	AD650S	† AD (2840)	50	TCA780	Siemens						
				TCA280A	Signetics						
				TDA1023	Signetics						
				TDA1024	Signetics			105			
				S63059	SiliconG						
				S63079	SiliconG						
				U106	Telefunken						
				U217	Telefunken						
				TA7606	Toshiba			110			
				Dual Over/Under-Voltage Detector							
				ICL7665	Intersil (3513)						
				Dual Transistors, Monolithic, Matched, See Linear-Arrays							

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

LINEAR Master Selection Guide





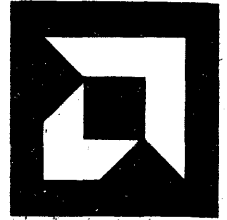
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# Advanced Micro Devices



## LINEAR Am6012

### Distinctive Characteristics

- All grades 12-bit monotonic over temperature
- Differential nonlinearity to  $\pm 0.012\%$  (13 bits) max. over temperature
- Trimless design is inherently monotonic
- Fast settling output current: 250nsec
- Full scale current 4mA
- High output impedance and compliance:  $-5$  to  $+10V$

- 100% MIL-STD-883 reliability assurance testing
- Differential current outputs
- Low cost
- High-speed multiplying capability
- Direct interface to TTL, CMOS, ECL, HTL, NMOS
- Performance unchanged over supply range
- Low power consumption: 230mW
- $R_{OUT}, C_{OUT}$  independent of logic code

### GENERAL DESCRIPTION

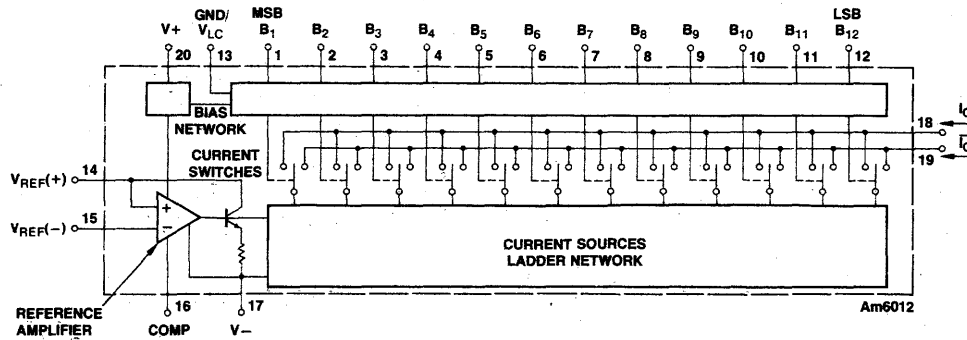
The Am6012 series of 12-bit monolithic multiplying Digital to Analog Converters represent a new level of high speed and accuracy coupled with low cost. The Am6012 is the first 12-bit D/A Converter ever built using standard processing without the requirements of thin film resistors and/or active trimming of individual devices. The Am6012 uses sophisticated new circuit design concepts that give inherent monotonicity without requiring ultra precision internal components.

The Am6012 design guarantees a more uniform step size than is possible with standard binarily weighted DAC's. This  $\pm 1/2$  LSB differential nonlinearity is desirable in many applications where local linearity is critical. The uniform step size allows finer resolution of levels and in most applications is more useful than conformance to an ideal straight line from zero to full scale.

The Am6012 has high voltage compliance, high impedance dual complementary outputs which increase its versatility and enable differential operation to effectively double the peak to peak output swing. These outputs can be used directly without op amps in many applications. The dual complementary outputs can also be connected in A/D converter applications to present a constant load current and significantly reduce switching transients and increase system throughput. Output full scale current is specified at 4mA, allowing use of smaller load resistors to minimize the output RC delay which usually dominates settling time at the 12-bit level.

The Am6012 series guarantees full 12-bit monotonicity for all grades and differential nonlinearity as tight as  $\pm 0.012\%$  (13 bits) over the entire temperature range. Device performance is essentially independent of power supply voltage. The devices work over a wide operating range of  $+5, -12$  volts to  $\pm 18$  volts.

### FUNCTIONAL DIAGRAM

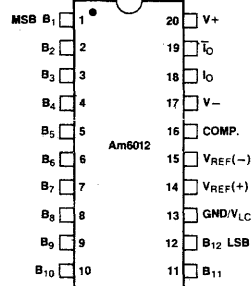


LIC-846

### ORDERING INFORMATION

Order Number	Package	Temperature Range	Differential Nonlinearity
AM6012ADM	Ceramic	$-55$ to $+125^{\circ}C$	$\pm 0.012\%$
AM6012DM	Ceramic	$-55$ to $+125^{\circ}C$	$\pm 0.025\%$
AM6012ADC	Ceramic	$0$ to $+70^{\circ}C$	$\pm 0.012\%$
AM6012APC	Plastic	$0$ to $+70^{\circ}C$	$\pm 0.012\%$
AM6012DC	Ceramic	$0$ to $+70^{\circ}C$	$\pm 0.025\%$
AM6012PC	Plastic	$0$ to $+70^{\circ}C$	$\pm 0.025\%$

### CONNECTION DIAGRAM — Top View

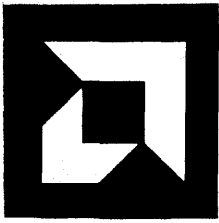


Note:  
Pin 1 is marked  
for orientation.

LIC-847

LINEAR

Advanced Micro Devices



# Advanced Micro Devices

## LINEAR Am6012

**MAXIMUM RATINGS** above which useful life may be impaired

Operating Temperature		Power Supply Voltage	±18V
Am6012ADM, Am6012DM	-55 to +125°C	Logic Inputs	-5 to +18V
Am6012ADC, Am6012DC	0 to +70°C	Analog Current Outputs	-8 to +12V
Am6012APC, Am6012PC	0 to +70°C	Reference Inputs V <sub>14</sub> , V <sub>15</sub>	V- to V+
Storage Temperature	-65 to +125°C	Reference Input Differential Voltage (V <sub>14</sub> to V <sub>15</sub> )	±18V
Lead Temperature (Soldering, 60 sec)	300°C	Reference Input Current (I <sub>14</sub> )	1.25mA

### ELECTRICAL CHARACTERISTICS

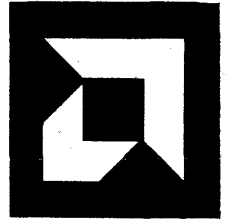
These specifications apply for V<sub>+</sub> = +15V, V<sub>-</sub> = -15V, I<sub>REF</sub> = 1.0mA, over the operating temperature range unless otherwise specified.

Parameter	Description	Test Conditions	Am6012A			Am6012			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
	Resolution		12	12	12	12	12	12	Bits
	Monotonicity		12	12	12	12	12	12	Bits
D.N.L.	Differential Nonlinearity	Deviation from ideal step size	-	-	±0.12	-	-	±0.25	%FS
			13	-	-	12	-	-	Bits
N.L.	Nonlinearity	Deviation from ideal straight line	-	-	±0.05	-	-	±0.05	%FS
I <sub>FS</sub>	Full Scale Current	V <sub>REF</sub> = 10.000V R <sub>14</sub> = R <sub>15</sub> = 10.000kΩ T <sub>A</sub> = 25°C	3.967	3.999	4.031	3.935	3.999	4.063	mA
TCI <sub>FS</sub>	Full Scale Tempco		-	±5	±20	-	±10	±40	ppm/°C
			-	±0.0005	±0.002	-	±0.001	±0.004	%FS/°C
V <sub>OC</sub>	Output Voltage Compliance	D.N.L. Specification guaranteed over compliance range R <sub>OUT</sub> > 10 megohms typ.	-5	-	+10	-5	-	+10	Volts
I <sub>FSS</sub>	Full Scale Symmetry	I <sub>FS</sub> - I <sub>FS</sub>	-	±0.2	±1.0	-	±0.4	±2.0	μA
I <sub>ZS</sub>	Zero Scale Current		-	-	0.10	-	-	0.10	μA
t <sub>S</sub>	Settling Time	To ±1/2 LSB, all bits ON or OFF, T <sub>A</sub> = 25°C	-	250	500	-	250	500	nsec
t <sub>PLH</sub> t <sub>PHL</sub>	Propagation Delay - all bits	50% to 50%	-	25	50	-	25	50	nsec
C <sub>OUT</sub>	Output Capacitance		-	20	-	-	20	-	pF
V <sub>IL</sub>	Logic Input Levels	Logic "0"	-	-	0.8	-	-	0.8	Volts
V <sub>IH</sub>		Logic "1"	2.0	-	-	2.0	-	-	
I <sub>IN</sub>	Logic Input Current	V <sub>IN</sub> = -5 to +18V	-	-	40	-	-	40	μA
V <sub>IS</sub>	Logic Input Swing	V <sub>-</sub> = -15V	-5	-	+18	-5	-	+18	Volts
I <sub>REF</sub>	Reference Current Range		0.2	1.0	1.1	0.2	1.0	1.1	mA
I <sub>15</sub>	Reference Bias Current		0	-0.5	-2.0	0	-0.5	-2.0	μA
dl/dt	Reference Input Slew Rate	R <sub>14(eq)</sub> = 800Ω CC = 0pF	4.0	8.0	-	4.0	8.0	-	mA/μs
PSSI <sub>FS+</sub> PSSI <sub>FS-</sub>	Power Supply Sensitivity	V <sub>+</sub> = +13.5V to +16.5V, V <sub>-</sub> = -15V V <sub>-</sub> = -13.5V to -16.5V, V <sub>+</sub> = +15V	-	±0.0005	±0.01	-	±0.0005	±0.01	%FS/%
V <sub>+</sub> V <sub>-</sub>	Power Supply Range	V <sub>OUT</sub> = 0V	4.5	-	18	4.5	-	18	Volts
			-18	-	-10.8	-18	-	-10.8	
I <sub>+</sub> I <sub>-</sub> I <sub>+</sub> I <sub>-</sub>	Power Supply Current	V <sub>+</sub> = +5V, V <sub>-</sub> = -15V V <sub>+</sub> = +15V, V <sub>-</sub> = -15V	-	5.7	8.5	-	5.7	8.5	mA
			-	-13.7	-18.0	-	-13.7	-18.0	
			-	5.7	8.5	-	5.7	8.5	
			-	-13.7	-18.0	-	-13.7	-18.0	
P <sub>D</sub>	Power Dissipation	V <sub>+</sub> = +5V, V <sub>-</sub> = -15V	-	234	312	-	234	312	mW
		V <sub>+</sub> = +15V, V <sub>-</sub> = -15V	-	291	397	-	291	397	

LINEAR

Advanced Micro Devices

# Advanced Micro Devices



## LINEAR Am6108 • Am6148

### DISTINCTIVE CHARACTERISTICS

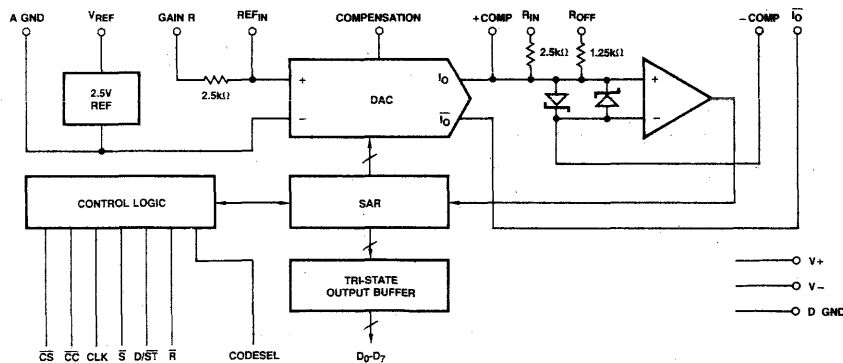
- 1  $\mu$ s conversion time
- Trimmed internal voltage reference
- 0.1% nonlinearity
- Ratiometric operation
- Low operating voltages
- Internal matched gain, reference and offset resistors
- Microprocessor compatible
- 3-state outputs
- Pin-programmable unipolar or bipolar two's complement conversion
- Conversion complete available as interrupt or as multiplexed output on data bus
- Available in slim, 24-pin, 0.3" and standard, 28-pin, 0.6" packages

### FUNCTIONAL DESCRIPTION

The Am6108 and Am6148 are microprocessor-compatible, 8-bit, high-speed, analog-to-digital converters. They include a precision reference, DAC, comparator, SAR, scale resistors, 3-state output buffers and control logic. The Am6108 is available in a standard .600-inch-wide, 28-pin package, and the Am6148 is offered in a space-saving, .300-inch-wide, 24-pin package. The Am6108/Am6148 are capable of completing an 8-bit conversion in under one microsecond and can handle input voltage ranges of 0 to +10V, 0 to +5V, and  $\pm 5$ V without external components. With appropriate external resistors, the user can program the device to operate on other input signal ranges (2 or 3 precision resistors are required). Full 8-bit monotonic performance with no missing codes is guaranteed over temperature. Both devices have 3-state outputs for bus compatibility and a status output.

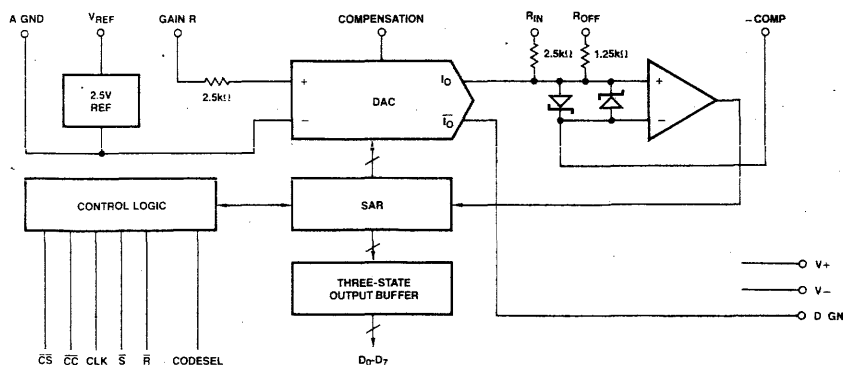
The Am6108/Am6148 are useful in microprocessor-based systems or can be used in a stand-alone mode. The conversion time is short enough to allow most microprocessors to accept data immediately after requesting a conversion. Applications include Analog I/O subsystems, process control and servo-control.

Am6108 EQUIVALENT CIRCUIT



ABI-092

Am6148 EQUIVALENT CIRCUIT



ABI-093

LINEAR

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## LINEAR Am6108 • Am6148

### MAXIMUM RATINGS

above which useful life may be impaired

V+ to D GND	-0.3 to +7.0V	Voltage at GAIN R, REF <sub>IN</sub>	V- to V+
V- to D GND	+0.3 to -7.0V	Voltage at R <sub>IN</sub> , R <sub>OFF</sub>	±12V
Max Differential V+ to V-	±12V	DAC Compliance Voltage	-2 to +12V
Digital Inputs to D GND	-0.5 to +6.0V	Operating Temperature	0 to +70°C
A GND to D GND	±1V	Storage Temperature	-65 to +150°C
V <sub>REF</sub> Max Output Current	15mA	Lead Temperature (Soldering 60 sec)	300°C
Max Input Current at REF <sub>IN</sub>	2mA	Minimum Operating Voltage	9.7V

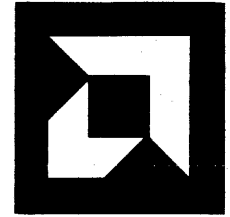
### ELECTRICAL CHARACTERISTICS

(These specifications apply for V<sup>+</sup> = +5V ± 5%, V<sup>-</sup> = -5.2V ± 5%, V<sub>REF</sub> connected to GAIN R, 0°C ≤ T<sub>A</sub> ≤ 70°C and f<sub>CLOCK</sub> = 500KHz)

Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Transfer Characteristics</b>						
	Resolution		8	8	8	Bits
	Monotonicity		8	8	8	Bits
	Differential Nonlinearity			±1/4	±1/2	LSB
	Linearity			±1/4	±1/2	LSB
	Inherent Quantization Error				±1/2	LSB
	Unipolar Gain Error	V <sub>IN</sub> = 0 to +5V		±1½	±4	LSB
		V <sub>IN a</sub> = 0 to +10V		±1	±2	
	Unipolar Offset Error			±1/2	±1	LSB
	Bipolar Gain Error	V <sub>IN</sub> = -5V to +5V		±1	±2	LSB
	Bipolar Offset Error			±3/4	±1½	LSB
	Positive Power Supply Sensitivity	V <sup>+</sup> = +5V ± 5%		0.02	0.2	%FS
	Negative Power Supply Sensitivity	V <sup>-</sup> = -5.2V ± 5%		0.02	0.2	%FS
<b>Internal Reference</b>						
V <sub>REF</sub>	Reference Voltage	I <sub>REF</sub> = 1mA	2.485	2.5	2.515	Volts
V <sub>REF</sub> /T <sub>A</sub>	Reference Voltage Tempco			20		ppm/°C
ΔV <sub>REF</sub> /V <sub>REF</sub>	Load Regulation	I <sub>REF</sub> = 1mA to 5mA		0.05	0.2	%V <sub>REF</sub>
ΔV <sub>REF</sub> /V <sub>REF</sub>	Line Regulation	V <sup>+</sup> = +5V ± 5%		0.05	0.2	%V <sub>REF</sub>
	Noise, f <sub>n</sub> = 10KHz to 1MHz			20		μV <sub>rms</sub>
<b>Analog Inputs</b>						
<b>Input Resistance</b>						
	±5V			2.5		KΩ
	0 to 10V			2.5		KΩ
	0 to 5V			1.25		KΩ
<b>Input Capacitance</b>						
	R <sub>IN</sub> , R <sub>OFF</sub> , REF <sub>IN</sub> *, GAIN R			2		pF
	I <sub>O</sub> *			20		pF
	+COMP*			20		pF
	-COMP			2		pF
<b>Digital Inputs</b>						
<b>Logic Level Input Voltage</b>						
V <sub>IH</sub>	Logic 1		2.0			Volts
V <sub>IL</sub>	Logic 0				0.8	Volts
<b>Logic Level Input Current</b>						
I <sub>IH</sub>	Logic 1	V <sub>IN</sub> = 2.7V			40	μA
I <sub>IL</sub>	Logic 0	V <sub>IN</sub> = 0.4V			10	μA

\*Function available on the Am6108 only.

# Advanced Micro Devices



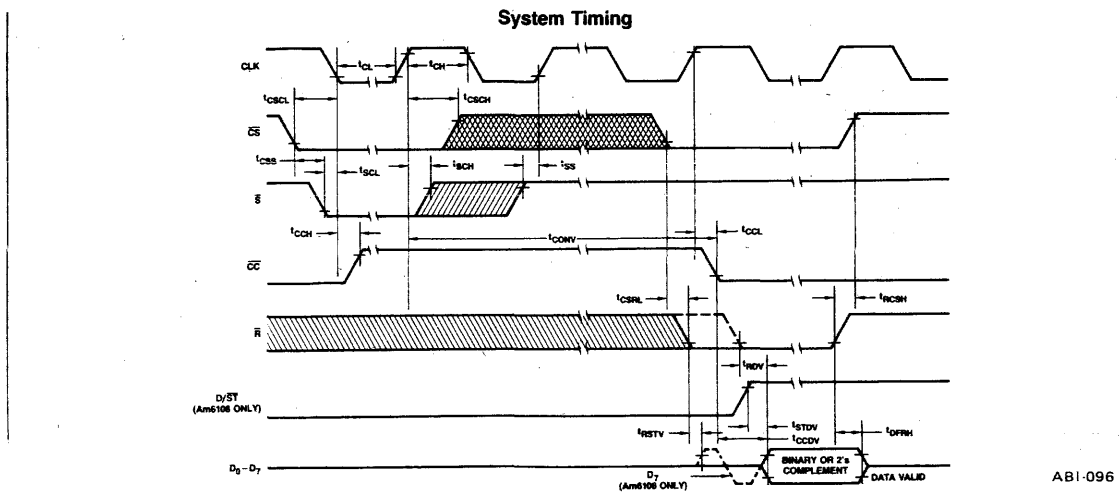
## LINEAR Am6108 • Am6148

### ELECTRICAL CHARACTERISTICS (Cont.)

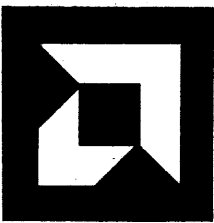
Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Digital Outputs</b>						
<b>Logic Level Output Voltages</b>						
$V_{OH}$	Logic 1	$I_{OH} = -400\mu A$	2.4			Volts
$V_{OL}$	Logic 0	$I_{OL} = 8mA$			0.5	Volts
$I_{SC}$	Output Short Circuit Current			-40		mA
<b>Off-State Output Current</b>						
		$V_O = 2.4V$		20		$\mu A$
		$V_O = 0.4V$		-20		$\mu A$
<b>Power Requirements</b>						
$I^+$	Positive Supply Current			44	60	mA
$I^-$	Negative Supply Current			-65	-85	mA
	Power Dissipation			600	800	mW

### SYSTEM TIMING

Parameters	Description	Min	Typ	Max	Units
$t_{CONV}$	Conversion Time		1	2	$\mu s$
$t_{CSS}$	$\overline{CS}$ Low to $\overline{S}$ Low	0			ns
$t_{CSCL}$	$\overline{CS}$ Low to CLK Low	0			ns
$t_{SCL}$	$\overline{S}$ Low to CLK Low	0			ns
$t_{CCH}$	$\overline{CC}$ High from CLK Low		35	40	ns
$t_{SCH}$	$\overline{S}$ High from CLK High	0			ns
$t_{SS}$	$\overline{S}$ High before CLK High	10			ns
$t_{CSCH}$	$\overline{CS}$ High from CLK High	0			ns
$t_{CCL}$	$\overline{CC}$ Low from CLK High	20	30	40	ns
$t_{CSRL}$	$\overline{CS}$ Low to $\overline{R}$ Low	0			ns
$t_{RSTV}$	$\overline{R}$ Low to Status Valid on $D_7$ (Am6108 Only)	15	30	40	ns
$t_{STDV}$	Status to Data Valid on $D_7$ (Am6108 Only)	15	30	40	ns
$t_{CCDV}$	$\overline{CC}$ Low to Data Valid	15	30	40	ns
$t_{RDV}$	$\overline{R}$ Low to Data Valid	15	30	40	ns
$t_{DFRH}$	Data Float from $\overline{R}$ High	20	30	40	ns
$t_{RCSH}$	$\overline{R}$ High to $\overline{CS}$ High	0			ns
$t_{CL}$	CLK Low	50			ns
$t_{CH}$	CLK High	50			ns



$\overline{R}$  function available on the Am6108 only.



# Advanced Micro Devices

## LINEAR Am6112 PRELIMINARY DATA

### DISTINCTIVE CHARACTERISTICS

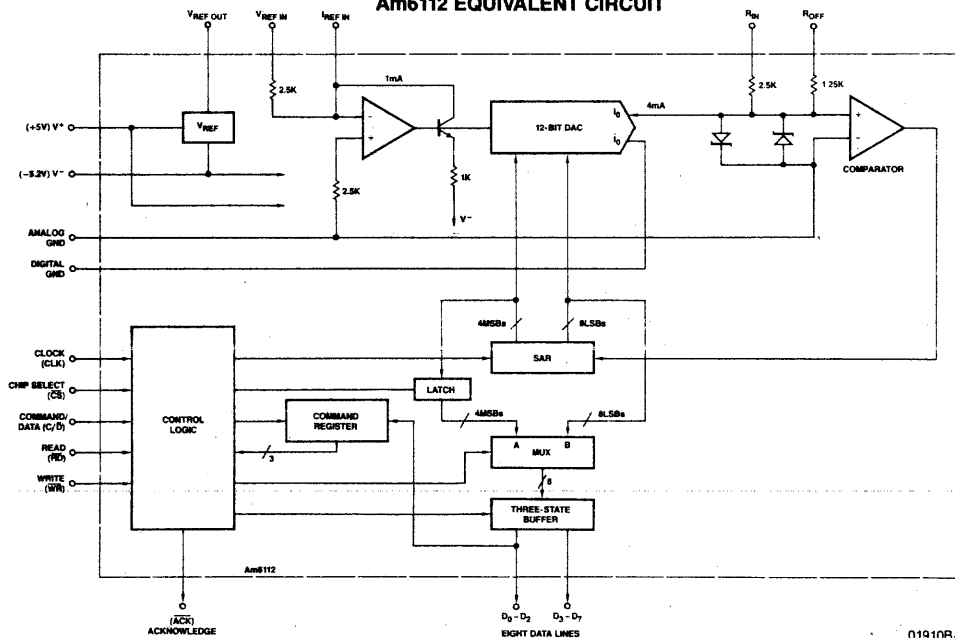
- First totally monolithic, high-speed 12-bit ADC
- 3  $\mu$ s typical conversion time
- Internal precision voltage reference
- No missing codes
- Easy interfacing with 8- and 16-bit microprocessors
- Internal command register for programmable modes of operation
- Offset binary or two's complement output code
- 0 to 10V, or  $\pm$  5V input range
- 24-pin package

### FUNCTIONAL DESCRIPTION

The Am6112 is the first fully monolithic microprocessor compatible 12-bit high-speed analog-to-digital converter. The Am6112 high-speed A/D contains a precision reference, DAC, comparator, SAR, scale resistors, output three-state buffers and comprehensive control logic, enabling the device to be interfaced with a variety of microprocessors. The Am6112 is capable of completing a 12-bit conversion in typically three microseconds and can handle input voltage ranges of 0 to 10V, and  $\pm$  5V without external components.

The Am6112 has four modes of microprocessor operation, and a stand-alone mode. These modes are software programmable, except for the stand-alone mode which is pin selectable. Applications include analog I/O subsystems, servo-control and high-speed digital signal processing of analog events.

### Am6112 EQUIVALENT CIRCUIT



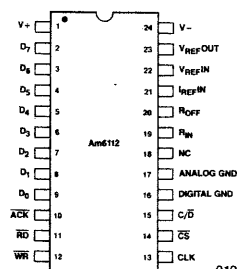
01910B-1

### ORDERING INFORMATION

Order the part number according to the table below to obtain the desired package, temperature range, and screening level.

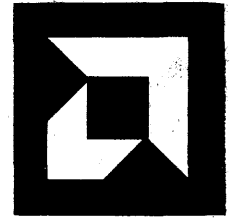
Order Number	Package Type	Temperature Range
AM6112DC	Hermetic DIP	0 to +70°C
AM6112XC	Dice	0 to +70°C
AM6112DM	Hermetic DIP	-55 to +125°C

### CONNECTION DIAGRAM — Top View D-24-1



01910B-2

Note: Pin 1 is marked for orientation.



## LINEAR Am6112

### ELECTRICAL CHARACTERISTICS (Cont.)

Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Digital Outputs</b>						
<b>Logic Level Output Voltages</b>						
V <sub>OH</sub>	Logic 1	I <sub>OH</sub> = -400μA	2.4			Volts
V <sub>OL</sub>	Logic 0	I <sub>OL</sub> = 8mA			0.5	Volts
I <sub>SC</sub>	Output Short Circuit Current			-40		mA
I <sub>OZ</sub>	Off-State Output Current					
		V <sub>O</sub> = 2.4V		20		μA
		V <sub>O</sub> = 0.4V		-20		μA
<b>Power Requirements</b>						
I <sup>+</sup>	Positive Supply Current			40	80	mA
I <sup>-</sup>	Negative Supply Current			-60	-80	mA
	Power Dissipation			500	800	mW

### SYSTEM TIMING

Parameters	Description	Min	Typ	Max	Units
t <sub>CONV</sub>	Conversion Time R <sub>OFF</sub> connected to AGND or V <sub>REF OUT</sub> (Bipolar)		3.3		μs
	R <sub>OFF</sub> Open (Unipolar)		10		μs
t <sub>CL</sub>	CLK Low		125		ns
t <sub>CH</sub>	CLK High		125		ns

### DEFINITION OF TERMS

**Resolution:** The number of possible analog input levels an A/D will resolve. Expressed as either the number of output bits, or 1 part in 2<sup>n</sup> where n is the number of bits.

**Monotonicity (Missing Codes):** Monotonicity is a property of the DAC within a successive approximation ADC. Each increment in the digital code to the DAC is accompanied by an analog output that is greater than or equal to that of the preceding code. Monotonicity of the DAC is a necessary requirement for a successive approximation ADC to have no missing codes.

**Differential Nonlinearity:** The deviation between the actual code width of an A/D from the ideal code width. The code width is defined as the range of analog input which produces a given digital output code. An ideal value of a code width is equivalent to FSR/2<sup>n</sup>, where n is the number of bits.

**Linearity:** The deviation of each individual code from an ideal straight line transfer curve between zero and full scale, with the straight line measured from the middle of each particular code.

**Inherent Quantization Error:** Quantization Error is a direct consequence of the resolution of the A/D. All analog voltages within a given range are represented by a single digital output code. There is, therefore, an inherent ±1/2 LSB conversion error even for a perfect A/D.

**Gain Error:** Defined as the difference between the analog input levels required to produce the first and the last digital output code transitions. Gain error is a measure of the deviation between the actual gain from the ideal gain of FS-2 LSB.

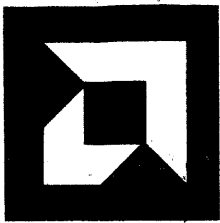
**Unipolar Offset Error:** Difference between the ideal (+1/2 LSB) and the actual analog input level required to produce the first digital code transition (00 . . . . 00 to 00 . . . . 01) over the complete temperature range.

**Bipolar Offset Error:** Difference between the ideal (1/2 FSR - 1/2 LSB) and the actual analog input level required to produce the major carry output digital code transition (from 01 . . . . 11 to 10 . . . . 00).

**Power Supply Sensitivity:** A measure of the change in gain and offset of the A/D resulting from a change in supply voltage. Usually expressed in total %FS for a percentage change in supply voltage.

**Conversion Time:** The measure of how long it takes for the A/D to arrive at the correct digital output code. It is the time between the clock edge that starts a conversion after receiving a start command and the edge of the status line (CC) which signifies that the conversion is completed.





# Advanced Micro Devices

## LINEAR Am6112

### MAXIMUM RATINGS above which useful life may be impaired

V+ to D GND	-0.3 to +7.0V	Voltage at R <sub>IN</sub> , R <sub>OFF</sub>	±12V
V- to D GND	+0.3 to -7.0V	Open Collector Current	20mA
Max Differential V+ to V-	±12V	Operating Temperature	0 to +70°C
Digital Inputs to D GND	-0.5 to +6.0V	Storage Temperature	-65 to +150°C
A GND to D GND	±1V	Lead Temperature (Soldering 60 sec)	300°C
V <sub>REF</sub> Max Output Current	15mA	Minimum Operating Voltage	9.7V
Max Input Current at REF <sub>IN</sub>	2mA	Max Package Dissipation	1W
Voltage at GAIN R, REF <sub>IN</sub>	V- to V+		

**ELECTRICAL CHARACTERISTICS** (These specifications apply for V<sup>+</sup> = +5V ± 5%, V<sup>-</sup> = -5.2V ± 5%, V<sub>REF</sub> connected per connection diagram, T<sub>A</sub> = 25°C, F<sub>Clock</sub> = 150kHz, 0 to +10V input range, R<sub>OFF</sub> open, stand-alone mode, unless otherwise stated)

Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Transfer Characteristics</b>						
	Resolution		12	12	12	Bits
	Monotonicity				12	Bits
	Differential Nonlinearity				±1	LSB
	Linearity			±1		LSB
	Inherent Quantization Error				±1/2	LSB
	Unipolar Gain Error			±2		LSB
	Unipolar Offset Error			±1/2		LSB
	Bipolar Gain Error	V <sub>IN</sub> = -5V to +5V		±4		LSB
	Bipolar Offset Error			±2		LSB
	Positive Power Supply Sensitivity	V <sup>+</sup> = +5V ± 5%		0.005		%FS
	Negative Power Supply Sensitivity	V <sup>-</sup> = -5.2V ± 5%		0.005		%FS
<b>Internal Reference</b>						
V <sub>REF</sub>	Reference Voltage	I <sub>REF</sub> = 1mA	2.490	2.5	2.510	Volts
V <sub>REF</sub> /T <sub>A</sub>	Reference Voltage Tempco			8		ppm/°C
ΔV <sub>REF</sub> /V <sub>REF</sub>	Load Regulation	I <sub>REF</sub> = 1mA to 5mA		0.005		%V <sub>REF</sub>
ΔV <sub>REF</sub> /V <sub>REF</sub>	Line Regulation	V <sup>+</sup> = +5V ± 5%, -5.2 ± 5%		0.005		%V <sub>REF</sub>
	Noise, N = 10kHz to 1MHz					μV <sub>RMS</sub>
<b>Analog Inputs</b>						
R <sub>IN</sub>	<b>Input Resistance</b>					
	±5V			2.5		KΩ
	0 to 10V			2.5		KΩ
C <sub>IN</sub>	<b>Input Capacitance</b>					
	R <sub>IN</sub> , R <sub>OFF</sub> , REF <sub>IN</sub> , GAIN R			2		pF
<b>Digital Inputs</b>						
<b>Logic Level Input Voltage</b>						
V <sub>IH</sub>	Logic 1		2.0			Volts
V <sub>IL</sub>	Logic 0				0.8	Volts
<b>Logic Level Input Current</b>						
I <sub>IH</sub>	Logic 1	V <sub>IN</sub> = 2.7V			40	μA
I <sub>IL</sub>	Logic 0	V <sub>IN</sub> = 0.4V			10	μA

# Advanced Micro Devices

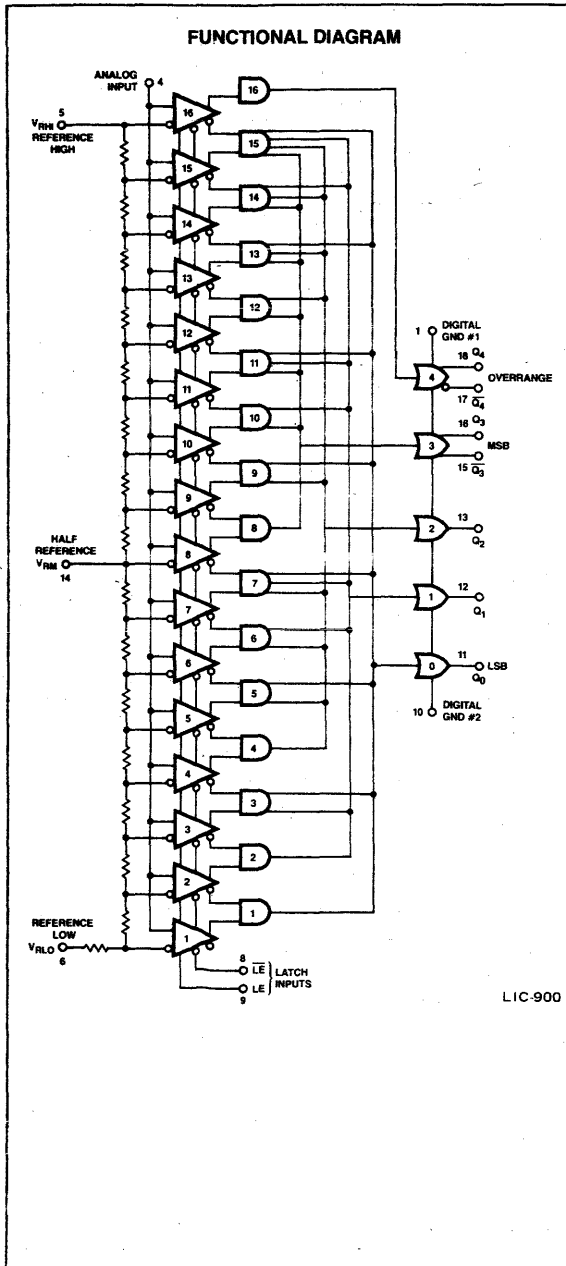


## LINEAR Am6688

### Distinctive Characteristics:

- 100MHz sampling rate
- 5ns maximum encode delay
- 4-bit resolution, expandable to 8 bits
- 8-bit accuracy

- Large bipolar input voltage range
- Low input current
- Q and  $\bar{Q}$  Outputs on MSB for 2's complement conversion
- 100% MIL-STD-883 reliability assurance testing



### FUNCTIONAL DESCRIPTION

The Am6688 4-bit quantizer consists of an array of 16 high-speed ECL sampling comparators, a resistor voltage divider, and an ECL-compatible binary encoder. It will accurately quantize an analog voltage into 15 equally-spaced levels and output a 4-bit binary digital word at sampling rates up to 100 MHz.

Resolution above 4-bits, up to a maximum of 8, may be obtained by stacking quantizers (n bits of resolution requires  $2^{n-4}$  quantizers). An overrange output signal is provided to indicate that the input signal has exceeded the full-scale limit. This overrange output is also the enable gating signal used to encode the higher-order bits of the output in a stacked configuration.

The high speed latch enable inputs are intended to be driven from the complementary outputs of a standard ECL gate or a high-speed comparator such as the Am685. If LE is driven high and  $\bar{LE}$  is driven low, the quantizer is in the sample mode and operates like a low-gain, high-bandwidth amplifier. When LE is driven low and  $\bar{LE}$  is driven high, the quantizer will hold its existing digital binary output word.

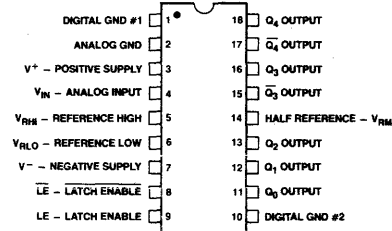
The outputs are open emitters, requiring external pull-down resistors of at least 200 $\Omega$  to -2V or 500 $\Omega$  to -5.2V.

These devices can be used in video data conversion and time-base correction, radar signal processing, nuclear pulse-height analysis, and other systems requiring very high-speed analog-to-digital conversion.

### ORDERING INFORMATION

Order Number	Temperature Range	Maximum Error
Am6688DL-8	-30 to +85°C	±5mV
Am6688DL-7	-30 to +85°C	±10mV
Am6688DL-6	-30 to +85°C	±20mV
Am6688DM-8	-55 to +125°C	±5mV
Am6688DM-7	-55 to +125°C	±10mV
Am6688DM-6	-55 to +125°C	±20mV

### CONNECTION DIAGRAM Top View



Note: Pin 1 is marked for orientation.

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LINEAR

Advanced Micro Devices



# Advanced Micro Devices

## LINEAR Am6688

### MAXIMUM RATINGS (Above which the useful life may be impaired)

Supply Voltage: Positive	+7V	Output Current (each output)	15mA
Negative	-6V	Temperature: Operating, Am6688DL	-30 to +85°C
Input Voltage: Analog	-5V to +3V	Am6688DM	-55 to +125°C
References	-5V to +3V	Storage Junction	-65 to +150°C
Digital	-5V to 0V	Lead (soldering, 60sec)	+175°C
Differential Voltage: Analog Input to References	±6V	Minimum Operating Voltage (V <sup>+</sup> to V <sup>-</sup> )	+300°C
Analog Gnd to Digital Gnds	±0.1V		

### ELECTRICAL CHARACTERISTICS OVER THE OPERATING TEMPERATURE RANGES

(Unless otherwise specified)

#### DC Characteristics

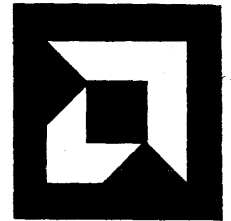
Symbol	Parameter (see definitions)	Conditions (Note 1)	Am6688DL		Am6688DM		Units
			Min	Max	Min	Max	
	Resolution		4		4		Bits
V <sub>OS</sub>	Error Voltage (each transition)	Am6688 - 8	-5	+5	-5	+5	mV
		- 7	-10	+10	-10	+10	mV
		- 6	-20	+20	-20	+20	mV
DNL	Differential Nonlinearity	ΔV <sub>REF</sub> = ΔV <sub>REF(min)</sub>	-0.5	0.5	-0.5	0.5	LSB
V <sub>IN</sub>	Input Voltage		-3.3	+2.7	-3.3	+2.7	V
ΔV <sub>REF</sub>	Reference Resistor Voltage (V <sub>RHI</sub> - V <sub>RLO</sub> )	Am6688 - 8	0.16	6.0	0.16	6.0	V
		- 7	0.32	6.0	0.32	6.0	V
		- 6	0.64	6.0	0.64	6.0	V
I <sub>REF</sub>	Reference Current	ΔV <sub>REF</sub> = 2.56V	6.0	17	5.0	18	mA
I <sub>B</sub>	Analog Input Current	V <sub>IN</sub> ≥ V <sub>RHI</sub>		230		250	μA
I <sub>L</sub>	Latch Input Current	V <sub>L</sub> ≥ V <sub>OH</sub>		200		220	μA
V <sub>OH</sub>	Output HIGH Voltage	T <sub>A</sub> = 25°C	-0.93	-0.72	-0.93	-0.72	V
		T <sub>A</sub> = T <sub>A(min)</sub>	-1.03	-0.80	-1.08	-0.83	V
		T <sub>A</sub> = T <sub>A(max)</sub>	-0.86	-0.64	-0.83	-0.58	V
V <sub>OL</sub>	Output LOW Voltage	T <sub>A</sub> = 25°C	-1.90	-1.62	-1.90	-1.62	V
		T <sub>A</sub> = T <sub>A(min)</sub>	-1.93	-1.65	-1.95	-1.66	V
		T <sub>A</sub> = T <sub>A(max)</sub>	-1.86	-1.58	-1.84	-1.54	V
I <sup>+</sup>	Positive Supply Current		100		100	mA	
I <sup>-</sup>	Negative Supply Current		-100		-100	mA	
P <sub>DISS</sub>	Power Dissipation	ΔV <sub>REF</sub> = 2.56V		1.2		1.2	W

#### Switching Characteristics (Note 2)

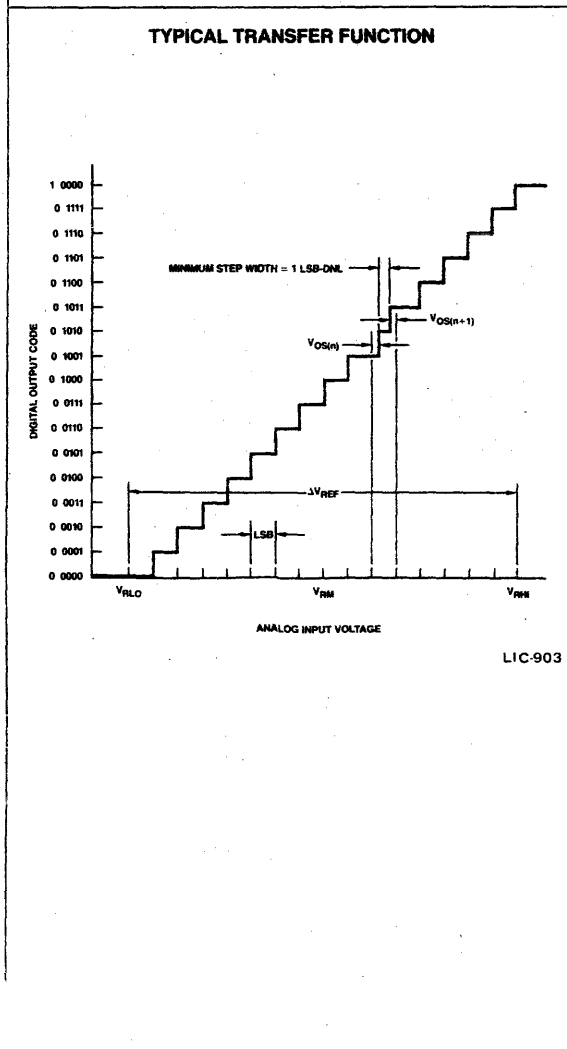
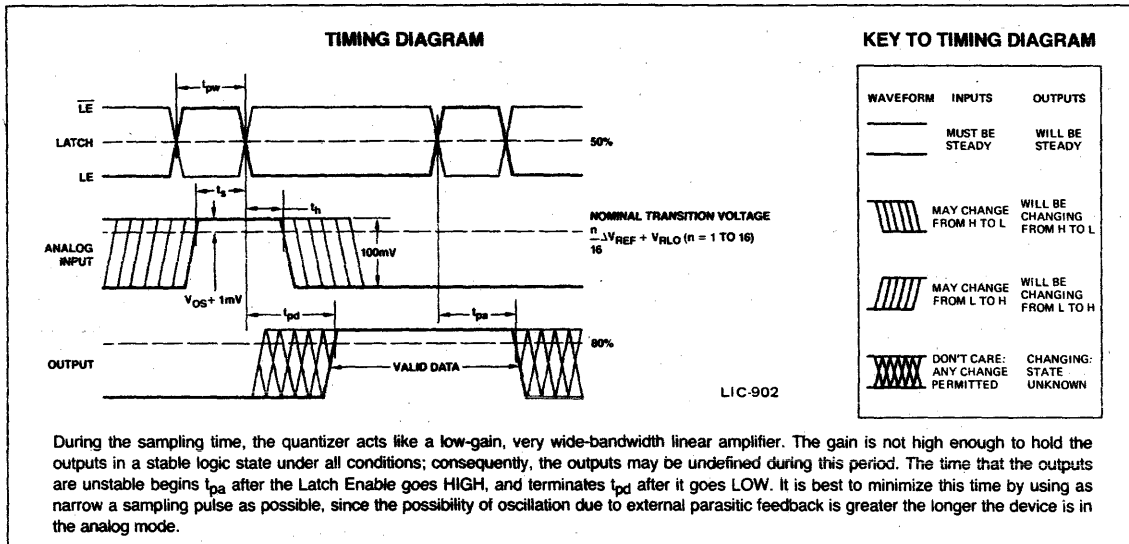
Symbol	Parameter	Conditions	Am6688DL	Am6688DM	Units
F <sub>MAX</sub>	Maximum Sampling Frequency		100	100	MHz
t <sub>pw</sub>	Minimum Sample Time		3	3	ns
t <sub>pd</sub>	Encode Delay (each transition)	T <sub>A(min)</sub> ≤ T <sub>A</sub> ≤ 25°C	5	5	ns
		T <sub>A</sub> = T <sub>A(max)</sub>	6	8	ns
t <sub>s</sub>	Minimum Set-up Time	T <sub>A</sub> = 25°C	3	3	ns
t <sub>h</sub>	Minimum Hold Time	T <sub>A</sub> = 25°C	1	1	ns
t <sub>pa</sub>	Analog Delay (each transition)	T <sub>A(min)</sub> ≤ T <sub>A</sub> ≤ 25°C	3	3	ns
		T <sub>A</sub> = T <sub>A(max)</sub>	4	4	ns

Notes: 1. Unless otherwise specified, V<sup>+</sup> = +6.0V, V<sup>-</sup> = -5.2V, V<sub>RHI</sub> = +2.56V, V<sub>RM</sub> = +1.28V, V<sub>RLO</sub> = 0V, t<sub>pw</sub> = 5ns, and R<sub>L</sub> = 100Ω to -2V at all outputs. The specifications given for V<sub>OS</sub>, DNL, and t<sub>pd</sub> apply over the full V<sub>IN</sub> range and for ±5% supply voltages. The Am6688 is designed to meet the specifications given in the table after thermal equilibrium has been established with a transverse air flow of 500 LFPM or greater.

2. Switching characteristics are for a 100mV analog input pulse level-shifted at each transition point to provide an overdrive of 1mV past the maximum specified error voltage.

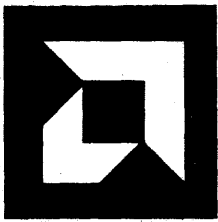


## LINEAR Am6688



### TRUTH TABLE

Analog Input (relative to $V_{RLO}$ )	$Q_4$	$Q_3$	$Q_2$	$Q_1$	$Q_0$
$V_{IN} < \frac{1}{16} \Delta V_{REF}$	L	L	L	L	L
$\frac{1}{16} \Delta V_{REF} < V_{IN} < \frac{1}{8} \Delta V_{REF}$	L	L	L	L	H
$\frac{1}{8} \Delta V_{REF} < V_{IN} < \frac{3}{16} \Delta V_{REF}$	L	L	L	H	L
$\frac{3}{16} \Delta V_{REF} < V_{IN} < \frac{1}{4} \Delta V_{REF}$	L	L	L	H	H
$\frac{1}{4} \Delta V_{REF} < V_{IN} < \frac{5}{16} \Delta V_{REF}$	L	L	H	L	L
$\frac{5}{16} \Delta V_{REF} < V_{IN} < \frac{3}{8} \Delta V_{REF}$	L	L	H	L	H
$\frac{3}{8} \Delta V_{REF} < V_{IN} < \frac{7}{16} \Delta V_{REF}$	L	L	H	H	L
$\frac{7}{16} \Delta V_{REF} < V_{IN} < \frac{1}{2} \Delta V_{REF}$	L	L	H	H	H
$\frac{1}{2} \Delta V_{REF} < V_{IN} < \frac{9}{16} \Delta V_{REF}$	L	H	L	L	L
$\frac{9}{16} \Delta V_{REF} < V_{IN} < \frac{5}{8} \Delta V_{REF}$	L	H	L	L	H
$\frac{5}{8} \Delta V_{REF} < V_{IN} < \frac{11}{16} \Delta V_{REF}$	L	H	L	H	L
$\frac{11}{16} \Delta V_{REF} < V_{IN} < \frac{3}{4} \Delta V_{REF}$	L	H	L	H	H
$\frac{3}{4} \Delta V_{REF} < V_{IN} < \frac{13}{16} \Delta V_{REF}$	L	H	H	L	L
$\frac{13}{16} \Delta V_{REF} < V_{IN} < \frac{7}{8} \Delta V_{REF}$	L	H	H	L	H
$\frac{7}{8} \Delta V_{REF} < V_{IN} < \frac{15}{16} \Delta V_{REF}$	L	H	H	H	L
$\frac{15}{16} \Delta V_{REF} < V_{IN} < \Delta V_{REF}$	L	H	H	H	H
$V_{IN} > \Delta V_{REF}$	H	L	L	L	L



# Advanced Micro Devices

## LINEAR Am6300

### DISTINCTIVE CHARACTERISTICS

- 2.5V  $\pm$ 0.25% temperature compensated reference
- Versatile 100mA output for driving external NPN or PNP power transistors
- Thermal shutdown
- Logic control power up enable
- Programmable delay and rise time for power supply
- $\pm$ 5% or  $\pm$ 10% over/under voltage detection/protection
- Programmable current limit detection/protection
- Programmable delays for the over/under voltage and current shutdown circuits
- Status outputs for fault conditions and output state

### GENERAL DESCRIPTION

The Am6300 Power Control Subsystem consists of a regulator section, an over/under voltage detection section, a current limit section and a reset and control section.

The regulator section contains a complete series pass voltage regulator with thermal shutdown, which uses external resistors to set the output voltage. Both the collector and emitter of the regulator output transistor are available to the user for flexibility in driving external power devices. The regulator also contains a precision, trimmed 2.5 volt reference which is capable of supplying 5mA of current for external purposes in addition to controlling the regulator and generating over and under voltage references.

The over/under voltage section compares the voltage at the sense input of the regulator to the internal reference and determines if the difference exceeds the user programmed limits,  $\pm$ 5% or  $\pm$ 10%. If one of the limits is exceeded for a period longer than the user programmed delay, the regulator shuts down and the voltage alarm output is activated. The regulator is reset by activating the power down inputs or removing power from the device.

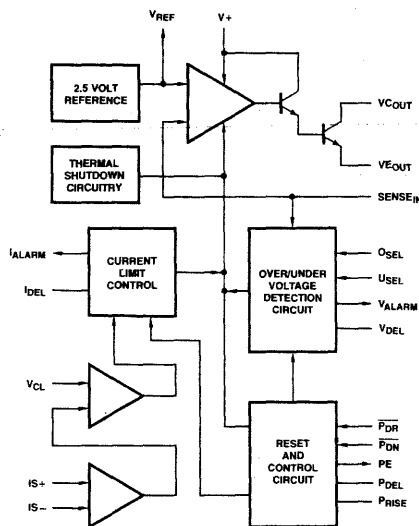
The current limit section detects overload current by means of an external sense resistor in series with  $V_{IN}$ , GND, or the output of

the regulator. The user programs the threshold of this detection circuit with external resistors. The regulator shuts down if a current overload is detected for a period longer than a user programmed delay and the current alarm is activated. The regulator is reset by activating the power down inputs or removing power from the device.

The reset and control section provides to the user, the ability to turn the regulator on and off by logic control. The start up of the regulator is delayed by a user programmed interval after the power up signal is received. After this delay the regulator output ramps up at a rate which is also determined by the user. When the output voltage levels off at the preset value, the over/under voltage and current limit circuits are activated and a power up output is activated which can be used to signal to the user that the supply is operating or enable other circuits.

The Am6300 allows the user a great deal of flexibility in power supply configuration and control. It can be operated locally or remotely in a stand alone configuration or with external power transistors to increase the output current. The Am6300 can be cascaded with other Am6300s for a sequenced supply application.

### BLOCK DIAGRAM

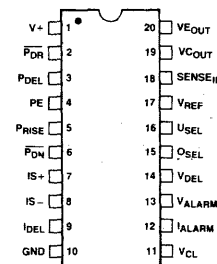


LIC-885

### ORDERING INFORMATION

Package Type	Temperature Range	Order Number
Hermetic DIP	-55 to +125°C	AM6300DM
Hermetic DIP	0 to 70°C	AM6300DC
Plastic DIP	0 to 70°C	AM6300PC

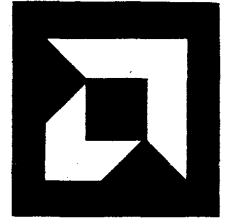
### CONNECTION DIAGRAM Top View



Pin 1 is marked for orientation.

LIC-886

# Advanced Micro Devices



## LINEAR Am6300

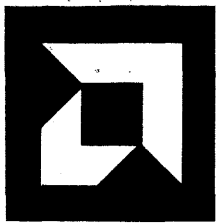
### MAXIMUM RATINGS

Pulse Voltage at V+ (50ms)	50V	Input Voltage (all pins)	Gnd to V+
Continuous Voltage at V+	40V	Maximum Output Current	100mA
Input-Output Voltage Differential	37.5V	Internal Power Dissipation	1000mW (Note 1)

### ELECTRICAL CHARACTERISTICS (Note 2)

Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Voltage Regulator</b>						
	Output Error (Note 3)		-1.0		+1.0	%V <sub>OUT</sub>
$\Delta V_{OUT}/\Delta V_{IN}$	Line Regulation, T <sub>A</sub> = 25°C	V <sub>IN</sub> = 12 to 40V	.02	.2		%V <sub>OUT</sub>
$\Delta V_{OUT}/\Delta I_L$	Load Regulation, T <sub>A</sub> = 25°C	I <sub>L</sub> = 1mA to 50mA	.03	.15		%V <sub>OUT</sub>
	Ripple Rejection	f = 1kHz T <sub>A</sub> = 25°C	74	86		dB
I <sub>BIAS</sub>	Sense Input Bias Current			2	4	μA
V <sub>IN</sub>	Input Operating Range		5		40	Volts
V <sub>OUT</sub>	Output Operating Range		2.5		37.5	Volts
V <sub>IN</sub> -V <sub>OUT</sub>	Input-Output Differential		2.5		37.5	Volts
I <sub>S</sub>	Supply Current			5	10	mA
<b>Voltage Reference</b>						
V <sub>REF</sub>	Reference Voltage	T <sub>A</sub> = 25°C	2.494	2.5	2.506	Volts
$\Delta V_{REF}/\Delta V_{IN}$	Line Regulation	V <sub>IN</sub> = 12 to 40V				%V <sub>REF</sub>
$\Delta V_{REF}/\Delta I_{REF}$	Load Regulation	I <sub>REF</sub> = 1mA to 5mA		.15	.30	%V <sub>REF</sub>
$\Delta V_{REF}/\Delta T_A$	Temperature Stability	0 to 70°C -55 to 125°C				ppm/°C
I <sub>SC</sub>	Short Circuit Current	V <sub>REF</sub> = 0V	10	30	60	mA
<b>Current Overload Circuit</b>						
V <sub>IS</sub> (diff)/V <sub>CL</sub>	Trip Point Ratio		.45	.5	.55	V/V
V <sub>CL</sub>	Trip Point Input Range		0		.4	Volts
V <sub>IS</sub> (diff)	Sense Voltage Input Range		0		.2	Volts
CMVR	Sense Input Common Mode Range		0		V <sub>IN</sub>	Volts
I <sub>BIAS</sub>	Input Bias Current (I <sub>S+</sub> , I <sub>S-</sub> )	V <sub>IS</sub> = 0V to +2V V <sub>IS</sub> = +2V to V <sub>IN</sub>		-8 +8	-20 +20	μA
I <sub>OS</sub>	Input Offset Current (I <sub>S+</sub> , I <sub>S-</sub> )			±1	±5	μA
I <sub>BIAS</sub>	Input Bias Current (V <sub>CL</sub> )			-25	-1.0	μA
<b>Voltage Protection Circuit</b>						
	+5% Error Trip Point	O <sub>SEL</sub> = 5V	4.5	5	5.5	%V <sub>OUT</sub>
	-5% Error Trip Point	U <sub>SEL</sub> = 5V	-4.5	-5	-5.5	%V <sub>OUT</sub>
	+10% Error Trip Point	O <sub>SEL</sub> = 0V	9	10	11	%V <sub>OUT</sub>
	-10% Error Trip Point	U <sub>SEL</sub> = 0V	-9	-10	-11	%V <sub>OUT</sub>
<b>Digital Characteristics (Note 4)</b>						
V <sub>IH</sub>	Input High Level		2.0		V <sub>IN</sub>	Volts
V <sub>IL</sub>	Input Low Level		0		.8	Volts
V <sub>OL</sub>	Open Collector Output Voltage	I <sub>OL</sub> = 8mA I <sub>OL</sub> = 15mA			.4 1.5	Volts

- Notes: 1. Power dissipation ratings apply for T<sub>A</sub> = 25°C. Derate linearly at 8mW/°C above 25°C for commercial parts and above 50°C for military parts.  
 2. All specifications are for V<sub>IN</sub> = 12V, V<sub>OUT</sub> = 5V, I<sub>OUT</sub> = 1mA, I<sub>REF</sub> = 1mA and over the operating temperature range unless otherwise specified.  
 3. Includes all errors associated with on chip reference source and temperature effects.  
 4. Digital Inputs are P<sub>DN</sub>, P<sub>DR</sub>, O<sub>SEL</sub>, U<sub>SEL</sub>  
 Digital Outputs are V<sub>ALARM</sub>, I<sub>ALARM</sub>, PE



# Advanced Micro Devices

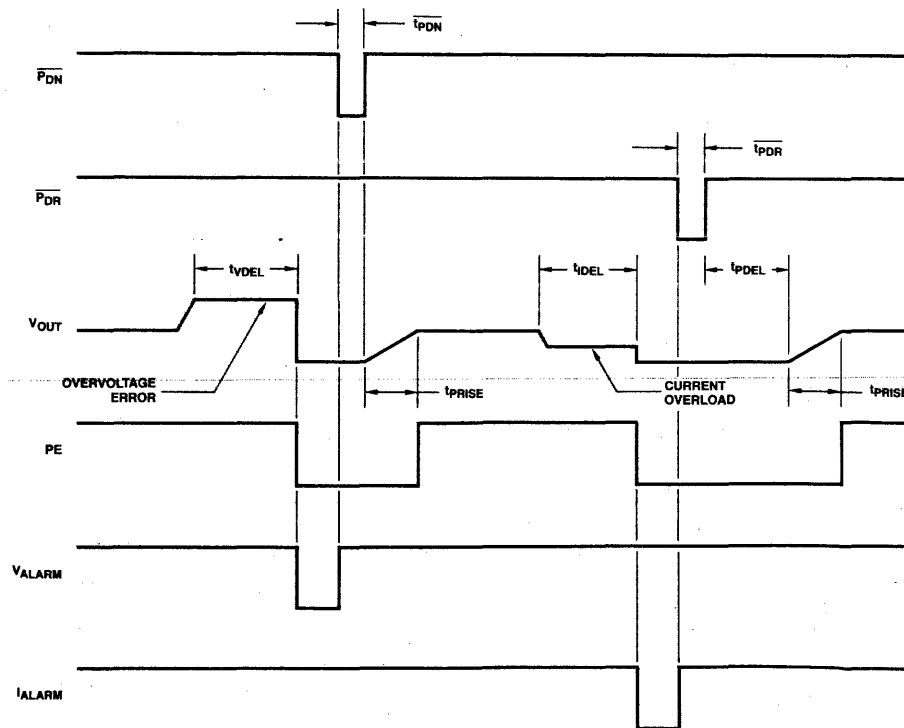
## LINEAR Am6300

### TIMING CHARACTERISTICS

Parameters	Description	Test Conditions	Min	Typ	Max	Units
$t_{IDEL}$ , $t_{VDEL}$	Current Overload and Voltage Error Power Down Delays	Without external cap		1.2		$\mu\text{s}$
		With external cap, additional delay		.22		$\mu\text{s/pF}$
$t_{PDEL}$	Power Up Delay (Note 5)	Without external cap		1.2		$\mu\text{s}$
		With external cap, additional delay		.22		$\mu\text{s/pF}$
$t_{PRISE}$	Power Up Rise Time (Note 5)	Without external cap		1.5		$\mu\text{s}$
		With external cap		.22		$\mu\text{s/pF}$
$t_{PDN}$ , $t_{PDR}$	Power Down Reset Pulse Width		10			$\mu\text{s}$

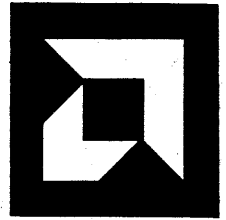
Note 5: It is necessary to make the total time  $t_{PDEL} + t_{PRISE}$  greater than the rise time of the supply to the Am6300 to insure proper power up.

### TIMING WAVEFORMS



LIC-887

# Advanced Micro Devices



## LINEAR Am6301

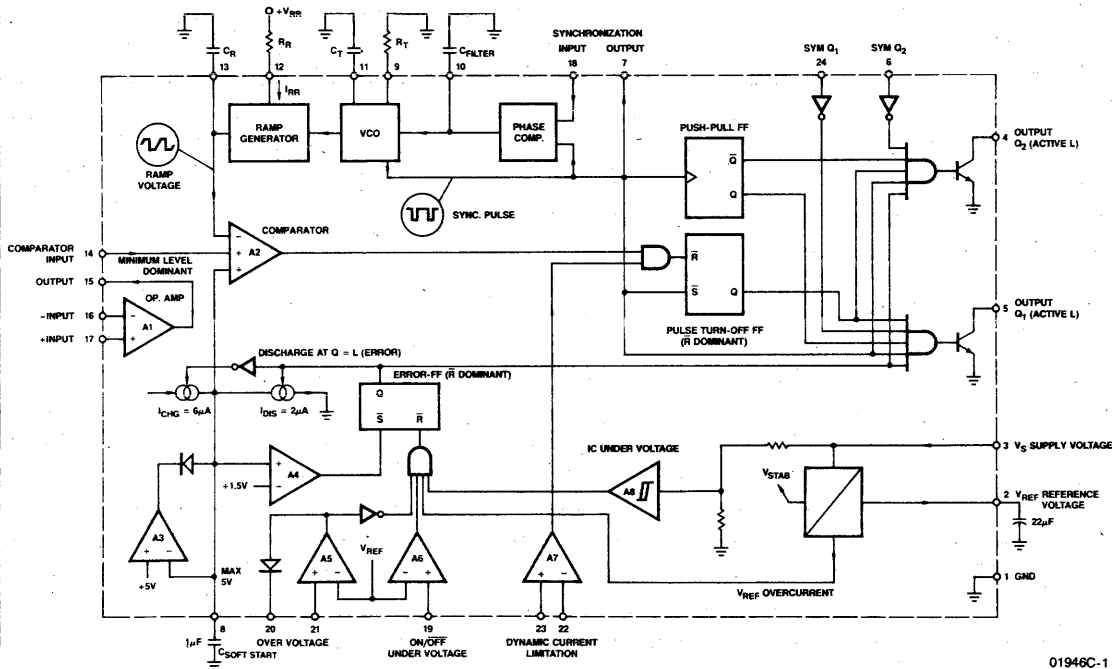
### DISTINCTIVE CHARACTERISTICS

- Feed-forward (line hum suppression)
- Output over/under voltage protection
- Input under voltage protection
- Cycle-by-cycle current limiting
- Soft start
- 250kHz max. oscillator frequency
- Phase lock capability
- 96% max. duty cycle
- Double pulse suppression
- Symmetry inputs for push-pull converter
- Remote shutdown
- Pin equivalent to the Siemens' TDA 4700

### GENERAL DESCRIPTION

The Am6301 Switching Power Supply Controller contains all the digital and analog functions necessary to control blocking, single-ended, or push-pull switching power supplies. It contains the voltage controlled oscillator, ramp generator, comparator, and reference for basic switched mode power supplies, as well as, a full complement of interface circuits and circuitry to protect both the power supply and its load.

### BLOCK DIAGRAM



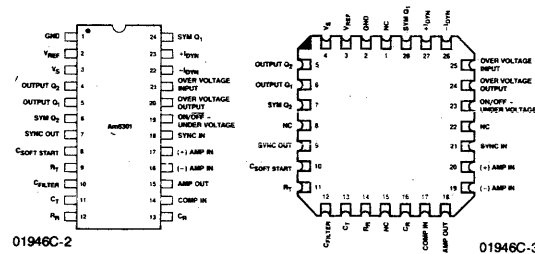
### ORDERING INFORMATION

Order the part number according to the table below to obtain the desired package, temperature range, and screening level.

Order Number	Package Type	Temperature Range
Am6301DM	Hermetic DIP	-55 to +125°C
Am6301LC	Leadless*	0 to 70°C
Am6301DL	Hermetic DIP	-25 to +85°C
Am6301DC	Hermetic DIP	0 to 70°C
Am6301PC	Plastic DIP	0 to 70°C

\*Availability of leadless packages will be announced.

### CONNECTION DIAGRAMS – Top Views

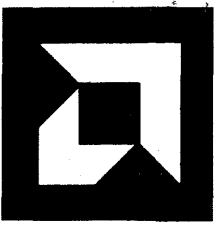


Note: Pin 1 is marked for orientation.

LINEAR

Advanced Micro Devices





# Advanced Micro Devices

## LINEAR Am6301

### Am6301DC, DL, PC, LC

**MAXIMUM RATINGS** (Above which useful life may be impaired)

Supply Voltage $V_S$	33V	Voltage at $Q_1, Q_2$	33V
Input Voltage (all inputs)	Gnd to $V_S$	Current at $Q_1, Q_2$	70mA
Output Voltage A5	33V	Input Current $R_{ramp}$	1mA
Output Voltage A1	6.5V	Operating Temperature	
Storage Temperature	-65 to +125°C	Am6301DL	-25 to +85°C
Lead Temperature (soldering 60 sec)	300°C	Am6301DC, Am6301PC, Am6301LC	0 to +70°C
Power Dissipation	(Note 3)		

### ELECTRICAL CHARACTERISTICS (Note 1)

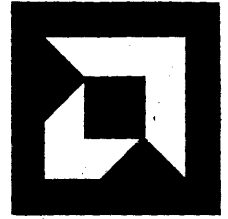
These specifications apply for  $11V \leq V_S \leq 30V$ ,  $f_{VCO} = 15kHz$ , over the operating temperature range unless otherwise specified.

Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Regulator</b>						
$V_S$	Supply Voltage		11		30	Volts
$I_S$	Supply Current	$C_T = 1nF, F = 100kHz$	8	12	20	mA
$f_{VCO}$	Operating Frequency Range		40		250K	Hz
<b>Reference</b>						
$V_{REF}$	Reference Voltage	$0 \leq I_{REF} \leq 5mA$	2.35	2.50	2.65	Volts
$\Delta V_{REF}$	Line Regulation	$V_S = 25V \pm 20\%$		5	15	mV
$\Delta V_{REF}$	Load Regulation	$I_{REF} = 0$ to 5mA		10	20	mV
$\Delta V_{REF}/\Delta T_A$	Temperature Stability			250	400	$\mu V/^\circ C$
<b>Oscillator</b>						
$f_{VCO}$	Operating Frequency Range		40		250K	Hz
$\Delta f/f$	Initial Tolerance	$\Delta C_T = 0\%, \Delta R_T = 0\%$			$\pm 7$	%
$\Delta f/f$	Frequency Stability	$V_S = 25V \pm 20\%$	-1		1	%
$t_{VCO}$	Fall Time	$C_T = 1nF$		1		$\mu s$
<b>Ramp Generator</b>						
$f_{ramp}$	Frequency Range		40		250K	Hz
$V_{ramp Hi}$	Voltage at $C_{ramp}$ High	$10\mu A \leq I_{ramp} \leq 400\mu A$		5.5		Volts
$V_{ramp Low}$	Voltage at $C_{ramp}$ Low	$10\mu A \leq I_{ramp} \leq 400\mu A$		1.8		Volts
$I_{ramp}$	Input Current at $R_{ramp}$		10		400	$\mu A$
<b>Synchronization</b>						
$V_{IH}$	Synchronization Input		2			Volts
$V_{IL}$					0.8	Volts
$V_{OH}$	Synchronization Output	$I_{OH} = 200\mu A$	4			Volts
$V_{OL}$		$I_{OL} = 1.6mA$			0.4	Volts
$I_B$	Input Bias Current		-5	-1		$\mu A$
<b>Comparator A2</b>						
$I_B$	Input Bias Current		-2	-1		$\mu A$
$t_{A2}$	Turn-Off Delay (Note 2)	$T_A = 25^\circ C$		470		ns
$V_{IN}$	Input Voltage for Duty Cycle	$T_{On}/T_{Off} = 0\%$		1.8		Volts
		$T_{On}/T_{Off} = Max$		5.0		Volts
<b>Soft Start</b>						
$I_{CHG}$	Charging Current			6		$\mu A$
$I_{DIS}$	Discharging Current			2		$\mu A$
$V_{LIM}$	Upper Limiting Voltage			5		Volts
$V_{TH}$	Reset Voltage			1.5		Volts

LINEAR

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## LINEAR Am6301

### ELECTRICAL CHARACTERISTICS (Cont.)

Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Operational Amplifier</b>						
AOL	Open Loop Gain		60	80		dB
VOS	Input Offset Voltage		-10		10	mV
$\Delta V_{OS}/\Delta T_A$	VOS Tempco		-30		30	$\mu V/^\circ C$
I <sub>B</sub>	Input Bias Current		-2	-0.5		$\mu A$
V <sub>CM</sub>	Common Mode Range		0		5	Volts
I <sub>O</sub>	Output Current		-3		1.5	mA
$\Delta V/\Delta t$	Output Slew Rate			1		V/ $\mu s$
BW	3dB Bandwidth			3		MHz
$\phi_1$	Phase Shift at 3MHz			120		degrees
VSW	Output Voltage Swing	$-3mA \leq I_L \leq 1.5mA$	1.5		5.5	Volts
<b>Symmetry</b>						
V <sub>IH</sub>	Input Voltage		2			Volts
V <sub>IL</sub>					0.8	Volts
I <sub>IL</sub>	Input Low Current		-2	-1		$\mu A$
<b>Output States Q<sub>1</sub>, Q<sub>2</sub></b>						
V <sub>OL</sub>	Output Voltage	I <sub>O</sub> = 20mA			1.1	Volts
I <sub>OH</sub>	Output Current	V <sub>OH</sub> = 30V			2	$\mu A$
<b>ON/OFF, Under Voltage</b>						
V <sub>TH</sub>	Threshold Voltage		V <sub>REF</sub> - 30mV	V <sub>REF</sub>	V <sub>REF</sub> + 30mV	Volts
I <sub>B</sub>	Input Bias Current		-2	-1		$\mu A$
t <sub>OFF</sub>	Turn-Off Delay			250		ns
t <sub>ERR</sub>	Error Recognition Time			50		ns
<b>Dynamic Current Limiting</b>						
V <sub>CM</sub>	Common Mode Range		0		4	Volts
V <sub>OS</sub>	Input Offset Voltage		-10		10	mV
I <sub>B</sub>	Input Bias Current		-2	-1		$\mu A$
t <sub>OFF</sub>	Turn-Off Delay			250		ns
t <sub>ERR</sub>	Error Recognition Time			50		ns
<b>Over Voltage</b>						
V <sub>TH</sub>	Threshold Voltage		V <sub>REF</sub> - 30mV	V <sub>REF</sub>	V <sub>REF</sub> + 30mV	Volts
I <sub>B</sub>	Input Bias Current		-2	-1		$\mu A$
I <sub>OH</sub>	Output Current	V <sub>OH</sub> = 5V	0		200	$\mu A$
t <sub>OFF</sub>	Turn-Off Delay			250		ns
t <sub>ERR</sub>	Error Recognition Time			50		ns
<b>Supply Under Voltage</b>						
V <sub>ON</sub>	Turn-On Threshold, V <sub>S</sub> Rising		8.8	9.6	10.5	Volts
V <sub>OFF</sub>	Turn-Off Threshold, V <sub>S</sub> Falling		8.5	9	10.0	Volts

Notes: 1. All typical values are specified at V<sub>S</sub> = 12V and T<sub>A</sub> = 25°C.

2. The A2 comparator turn-off delay is measured from pin 14 to pin 4 or pin 5. The input signal on pin 14 is a negative-going pulse from 3V to 0.5V. The delay is measured from the 2V level on the input to the 2V level on the output. The output current is set at 4mA. The signal on pin 14 goes negative when the ramp voltage is equal to 2V.

3. For T<sub>A</sub> greater than 25°C, derate to limit T<sub>J</sub> to a maximum of 150°C. For P<sub>D</sub> less than 750mW use typical thermal resistance as follows:

Typ	D-24-1	P-24-1	L-28-1	
$\theta_{JA}$	50	120	100	$^\circ C/watt$
$\theta_{JC}$	15	60	40	$^\circ C/watt$



# Advanced Micro Devices

## LINEAR Am6301

### Am6301 FUNCTIONAL DESCRIPTION

#### VOLTAGE-CONTROLLED OSCILLATOR (VCO)

The VCO (voltage-controlled oscillator) generates a sawtooth voltage at  $C_T$ . The duration of the falling edge is determined by the selection of  $C_T$ . The duration of the rising edge and thus the oscillator frequency is determined by  $R_T$  and  $C_T$ . Maximum oscillator frequency is 250kHz. The oscillator frequency can be varied for frequency synchronization purposes by varying the voltage at  $C_{filter}$ . The falling edge of the VCO generates the synchronization pulse and triggers the ramp generator and other parts of the Am6301.

#### RAMP GENERATOR – FEED-FORWARD CONTROL

The ramp generator is triggered by the synchronization pulse of the VCO and oscillates at the same frequency. The duration of the falling edge of the ramp generator must be shorter than the fall time of the VCO. The voltage of the rising edge of the ramp generator and a DC voltage at comparator A2 are compared for pulse width control of the output. The slope of the rising edge is adjusted via the current through  $R_R$ . This enables an additional superimposed control of the duty cycle dependent on the input voltage of the Switched Mode Power Supplies (SMPS). This capability (feed-forward) allows for compensation of a known interference (e.g. line hum).

#### PHASE COMPARATOR – SYNCHRONIZATION

If the Am6301 is operated without external synchronization, the synchronization input must be connected to the synchronization output, so that the phase comparator sets the voltage at  $C_{filter}$ . The VCO then oscillates at the frequency set by  $R_T$  and  $C_T$ . Other circuits can be synchronized with the synchronization output. The Am6301 can be synchronized to an external signal of any duty cycle. The synchronization input and output are TTL compatible.

#### PUSH-PULL FLIP-FLOP

The push-pull flip-flop is toggled by the falling edge of the VCO. This guarantees that only one of the two push-pull outputs can be enabled at any one time.

#### COMPARATOR A2 – PULSE WIDTH MODULATION

The two noninverting inputs of the comparator are switched in such a manner that the lowest level is always compared with the inverting input. As soon as the voltage of the rising sawtooth at  $C_R$  exceeds the lower of the two levels, both outputs are disabled via the pulse turn-off flip-flop.

#### REGULATING AMPLIFIER A1

A1 is a high-quality regulating amplifier. It can be used in the control loop to transmit the amplified error voltage onto the free noninverting input of the comparator A2. A voltage change is thus transformed into a duty cycle change. The common mode range of A1 covers 0 to +5V. A1's low output impedance allows the use of feedback for the adjustment of the regulator loop-characteristics.

#### PULSE TURN-OFF FLIP-FLOP

This flip-flop enables the outputs at the beginning of each half period, and upon an error signal from A7 or a turn-off signal from A2 switches the outputs off for the remainder of the half period. Double pulses at the output cannot occur.

#### COMPARATOR A3

A3 limits the voltage at the  $C_{soft\ start}$  pin (and also one input of A2) to a maximum of 5V. For a specified slope of the rising ramp generator edge, the duty cycle can be limited to a maximum value.

#### COMPARATOR A4

Comparator A4 has its switching threshold set to 1.5V and its output connected to the error flip-flop, so that when the voltage at capacitor  $C_{soft\ start}$  is less than 1.5V the flip-flop is set. The error flip-flop only accepts the set pulse if no reset signal is present. Thus, an output turn-on is prevented as long as an error signal is present.

#### SOFT START

The output duty cycle is a function of the lower of the two voltages at the noninverting inputs of A2. At the time the Am6301 is turned on, the voltage at capacitor  $C_{soft\ start}$  is equal to 0V. As long as no error exists, this capacitor is charged with a current of  $6\mu A$  to the maximum value of 5V. In the case of an error,  $C_{soft\ start}$  is discharged with a current of  $2\mu A$ . The error flip-flop is set when the  $C_{soft\ start}$  voltage is below 1.5V and the outputs are enabled if a reset signal is not present at the same time. The minimum ramp generator voltage is 1.8V, therefore, the soft start circuit only controls the duty cycle after the voltage at  $C_{soft\ start}$  exceeds 1.8V.

#### ERROR FLIP-FLOP

Error signals to input  $\bar{R}$  of the error flip-flop cause the outputs to be disabled immediately. The system turns on again using the soft start, after the error has been eliminated.

#### COMPARATOR A5 – OVER VOLTAGE

The input or output voltages of an SMPS can be monitored using A5. In the case of an over voltage, the error flip-flop immediately disables the IC outputs. After the over voltage is reduced, the SMPS turns back on using the soft start. The output of A5 can be fed back to the input. This causes the IC output stage to remain disabled even after elimination of the over voltage, until the supply voltage is briefly turned off, or the over voltage input is briefly connected to ground. To use this SCR-type action, the voltage to be monitored must be coupled resistively ( $\geq 5K\Omega$ ) to the over voltage comparator.

#### COMPARATOR A6 – ON/OFF UNDER VOLTAGE

The comparator A6 reacts to an under voltage relative to  $V_{REF}$  and switches the IC outputs off. The input voltage of the SMPS can for example, be monitored, turning the outputs off if the input voltage is below a desired level. When the input voltage returns to the desired level, the Am6301 turns back on using the soft start. This input can also be used as a TTL compatible ON/OFF control.

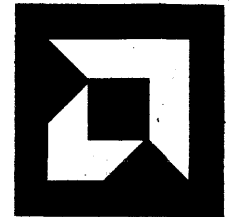
#### COMPARATOR A7 – DYNAMIC CURRENT LIMITING CIRCUIT

A7 provides for the recognition of over current in the switching transistors. The system is turned on again at the beginning of the half period after the error is eliminated, the soft start is not used however. The A7 common mode range extends from 0 to +4V so that the smallest voltage drops can be recognized. The delay time from the occurrence of an error to the disabling of the outputs is only 250ns.

#### COMPARATOR A8 – IC UNDER VOLTAGE

Comparator A8 prevents undefined operating conditions of the IC outputs if the IC supply voltage becomes too low. If  $V_S \leq 9V$ , the output stage is disabled. This condition is maintained until  $V_S = 0V$ . Built-in hysteresis prevents permanent switching at the comparator's switching threshold. At a supply voltage of  $V_S \geq 9.6V$ , the Am6301 turns on using the soft start.

# Advanced Micro Devices



## LINEAR Am6301

### ELECTRICAL CHARACTERISTICS (Cont.)

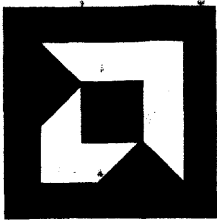
Parameters	Description	Test Conditions	Min	Typ	Max	Units
<b>Operational Amplifier</b>						
A <sub>OL</sub>	Open Loop Gain		60	80		dB
V <sub>OS</sub>	Input Offset Voltage		-10		10	mV
ΔV <sub>OS</sub> /ΔT <sub>A</sub>	V <sub>OS</sub> Tempco		-30		30	μV/°C
I <sub>B</sub>	Input Bias Current		-2	-0.5		μA
V <sub>CM</sub>	Common Mode Range		0		5	Volts
I <sub>O</sub>	Output Current		-3		1.5	mA
ΔV/Δt	Output Slew Rate			1		V/μs
BW	3dB Bandwidth			3		MHz
φ <sub>t</sub>	Phase Shift at 3MHz			120		degrees
V <sub>SW</sub>	Output Voltage Swing	-3mA ≤ I <sub>L</sub> ≤ 1.5mA	1.5		5.5	Volts
<b>Symmetry</b>						
V <sub>IH</sub>	Input Voltage		2			Volts
V <sub>IL</sub>					0.8	Volts
I <sub>IL</sub>	Input Low Current		-2	-1		μA
<b>Output States Q<sub>1</sub>, Q<sub>2</sub></b>						
V <sub>OL</sub>	Output Voltage	I <sub>O</sub> = 20mA			1.1	Volts
I <sub>OH</sub>	Output Current	V <sub>OH</sub> = 30V			2	μA
<b>ON/OFF, Under Voltage</b>						
V <sub>TH</sub>	Threshold Voltage		V <sub>REF</sub> - 30mV	V <sub>REF</sub>	V <sub>REF</sub> + 30mV	Volts
I <sub>B</sub>	Input Bias Current		-2	-1		μA
t <sub>OFF</sub>	Turn-Off Delay			250		ns
t <sub>ERR</sub>	Error Recognition Time			50		ns
<b>Dynamic Current Limiting</b>						
V <sub>CM</sub>	Common Mode Range		0		4	Volts
V <sub>OS</sub>	Input Offset Voltage		-10		10	mV
I <sub>B</sub>	Input Bias Current		-2	-1		μA
t <sub>OFF</sub>	Turn-Off Delay			250		ns
t <sub>ERR</sub>	Error Recognition Time			50		ns
<b>Over Voltage</b>						
V <sub>TH</sub>	Threshold Voltage		V <sub>REF</sub> - 30mV	V <sub>REF</sub>	V <sub>REF</sub> + 30mV	Volts
I <sub>B</sub>	Input Bias Current		-2	-1		μA
I <sub>OH</sub>	Output Current	V <sub>OH</sub> = 5V	0		200	μA
t <sub>OFF</sub>	Turn-Off Delay			250		ns
t <sub>ERR</sub>	Error Recognition Time			50		ns
<b>Supply Under Voltage</b>						
V <sub>ON</sub>	Turn-On Threshold, V <sub>S</sub> Rising		8.8	9.6	10.5	Volts
V <sub>OFF</sub>	Turn-Off Threshold, V <sub>S</sub> Falling		8.5	9	10.0	Volts

Notes: 1. All typical values are specified at V<sub>S</sub> = 12V and T<sub>A</sub> = 25°C.

2. The A2 comparator turn-off delay is measured from pin 14 to pin 4 or pin 5. The input signal on pin 14 is a negative-going pulse from 3V to 0.5V. The delay is measured from the 2V level on the input to the 2V level on the output. The output current is set at 4mA. The signal on pin 14 goes negative when the ramp voltage is equal to 2V.

3. For T<sub>A</sub> greater than 25°C, derate to limit T<sub>J</sub> to a maximum of 150°C. For P<sub>D</sub> less than 750mW use typical thermal resistance as follows:

Typ	D-24-1	P-24-1	L-28-1	
θ <sub>JA</sub>	50	120	100	°C/watt
θ <sub>JC</sub>	15	60	40	°C/watt



# Advanced Micro Devices

LINEAR  
Am6301

## Am6301 FUNCTIONAL DESCRIPTION

### VOLTAGE-CONTROLLED OSCILLATOR (VCO)

The VCO (voltage-controlled oscillator) generates a sawtooth voltage at  $C_T$ . The duration of the falling edge is determined by the selection of  $C_T$ . The duration of the rising edge and thus the oscillator frequency is determined by  $R_T$  and  $C_T$ . Maximum oscillator frequency is 250kHz. The oscillator frequency can be varied for frequency synchronization purposes by varying the voltage at  $C_{filter}$ . The falling edge of the VCO generates the synchronization pulse and triggers the ramp generator and other parts of the Am6301.

### RAMP GENERATOR – FEED-FORWARD CONTROL

The ramp generator is triggered by the synchronization pulse of the VCO and oscillates at the same frequency. The duration of the falling edge of the ramp generator must be shorter than the fall time of the VCO. The voltage of the rising edge of the ramp generator and a DC voltage at comparator A2 are compared for pulse width control of the output. The slope of the rising edge is adjusted via the current through  $R_R$ . This enables an additional superimposed control of the duty cycle dependent on the input voltage of the Switched Mode Power Supplies (SMPS). This capability (feed-forward) allows for compensation of a known interference (e.g. line hum).

### PHASE COMPARATOR – SYNCHRONIZATION

If the Am6301 is operated without external synchronization, the synchronization input must be connected to the synchronization output, so that the phase comparator sets the voltage at  $C_{filter}$ . The VCO then oscillates at the frequency set by  $R_T$  and  $C_T$ . Other circuits can be synchronized with the synchronization output. The Am6301 can be synchronized to an external signal of any duty cycle. The synchronization input and output are TTL compatible.

### PUSH-PULL FLIP-FLOP

The push-pull flip-flop is toggled by the falling edge of the VCO. This guarantees that only one of the two push-pull outputs can be enabled at any one time.

### COMPARATOR A2 – PULSE WIDTH MODULATION

The two noninverting inputs of the comparator are switched in such a manner that the lowest level is always compared with the inverting input. As soon as the voltage of the rising sawtooth at  $C_T$  exceeds the lower of the two levels, both outputs are disabled via the pulse turn-off flip-flop.

### REGULATING AMPLIFIER A1

A1 is a high-quality regulating amplifier. It can be used in the control loop to transmit the amplified error voltage onto the free noninverting input of the comparator A2. A voltage change is thus transformed into a duty cycle change. The common mode range of A1 covers 0 to +5V. A1's low output impedance allows the use of feedback for the adjustment of the regulator loop-characteristics.

### PULSE TURN-OFF FLIP-FLOP

This flip-flop enables the outputs at the beginning of each half period, and upon an error signal from A7 or a turn-off signal from A2 switches the outputs off for the remainder of the half period. Double pulses at the output cannot occur.

### COMPARATOR A3

A3 limits the voltage at the  $C_{soft\ start}$  pin (and also one input of A2) to a maximum of 5V. For a specified slope of the rising ramp generator edge, the duty cycle can be limited to a maximum value.

### COMPARATOR A4

Comparator A4 has its switching threshold set to 1.5V and its output connected to the error flip-flop, so that when the voltage at capacitor  $C_{soft\ start}$  is less than 1.5V the flip-flop is set. The error flip-flop only accepts the set pulse if no reset signal is present. Thus, an output turn-on is prevented as long as an error signal is present.

### SOFT START

The output duty cycle is a function of the lower of the two voltages at the noninverting inputs of A2. At the time the Am6301 is turned on, the voltage at capacitor  $C_{soft\ start}$  is equal to 0V. As long as no error exists, this capacitor is charged with a current of  $6\mu A$  to the maximum value of 5V. In the case of an error,  $C_{soft\ start}$  is discharged with a current of  $2\mu A$ . The error flip-flop is set when the  $C_{soft\ start}$  voltage is below 1.5V and the outputs are enabled if a reset signal is not present at the same time. The minimum ramp generator voltage is 1.8V, therefore, the soft start circuit only controls the duty cycle after the voltage at  $C_{soft\ start}$  exceeds 1.8V.

### ERROR FLIP-FLOP

Error signals to input  $\bar{R}$  of the error flip-flop cause the outputs to be disabled immediately. The system turns on again using the soft start, after the error has been eliminated.

### COMPARATOR A5 – OVER VOLTAGE

The input or output voltages of an SMPS can be monitored using A5. In the case of an over voltage, the error flip-flop immediately disables the IC outputs. After the over voltage is reduced, the SMPS turns back on using the soft start. The output of A5 can be fed back to the input. This causes the IC output stage to remain disabled even after elimination of the over voltage, until the supply voltage is briefly turned off, or the over voltage input is briefly connected to ground. To use this SCR-type action, the voltage to be monitored must be coupled resistively ( $\geq 5K\Omega$ ) to the over voltage comparator.

### COMPARATOR A6 – ON/OFF UNDER VOLTAGE

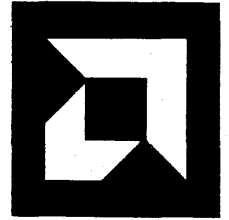
The comparator A6 reacts to an under voltage relative to  $V_{REF}$  and switches the IC outputs off. The input voltage of the SMPS can for example, be monitored, turning the outputs off if the input voltage is below a desired level. When the input voltage returns to the desired level, the Am6301 turns back on using the soft start. This input can also be used as a TTL compatible ON/OFF control.

### COMPARATOR A7 – DYNAMIC CURRENT LIMITING CIRCUIT

A7 provides for the recognition of over current in the switching transistors. The system is turned on again at the beginning of the half period after the error is eliminated, the soft start is not used however. The A7 common mode range extends from 0 to +4V so that the smallest voltage drops can be recognized. The delay time from the occurrence of an error to the disabling of the outputs is only 250ns.

### COMPARATOR A8 – IC UNDER VOLTAGE

Comparator A8 prevents undefined operating conditions of the IC outputs if the IC supply voltage becomes too low. If  $V_S \leq 9V$ , the output stage is disabled. This condition is maintained until  $V_S = 0V$ . Built-in hysteresis prevents permanent switching at the comparator's switching threshold. At a supply voltage of  $V_S \geq 9.6V$ , the Am6301 turns on using the soft start.



## LINEAR Am6301

### SYMMETRY

Saturation of the transformer core must be prevented in push-pull converters. The degree of saturation of the transformer can be determined with an external circuit; and, in relation to this, the on times of the outputs can be asymmetrically shortened. If the symmetry correction circuit is not required, the symmetry inputs must be connected to ground. The input levels are TTL compatible.

### OUTPUTS, DEADTIME

The two outputs  $Q_1$  and  $Q_2$  are transistors with open collectors. Their saturation voltage is 1V at 25mA. They operate in a push-pull mode and can be connected in parallel to drive single-ended converters with a maximum duty cycle of 96%. The time, during which only one of the two outputs is on, can be varied. The duration of the falling edge at the VCO is the same as the minimum time (dead time) during which both outputs are disabled simultaneously. The dead time, in push-pull SMPS, prevents the power transistors from being on at the same time.

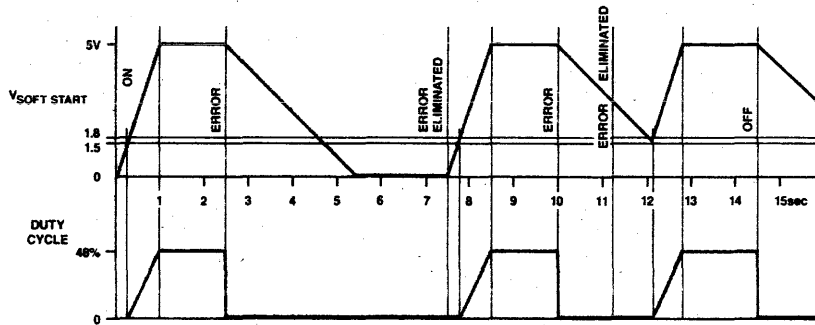


Figure 1. Am6301 System Timing Showing Soft Start, Error Timing, and Remote Shutdown. ( $C_{soft\ start} = 1\mu F$ )

ABI-021

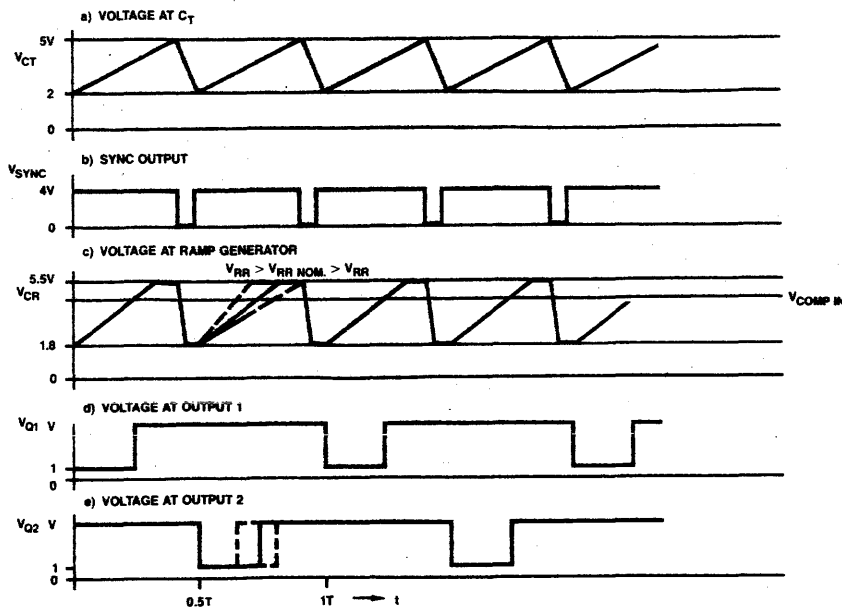
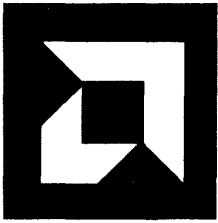


Figure 2. Am6301 Pulse Timing Diagram, Dotted Lines Show the Effect of Feed-Forward Control.

ABI-022



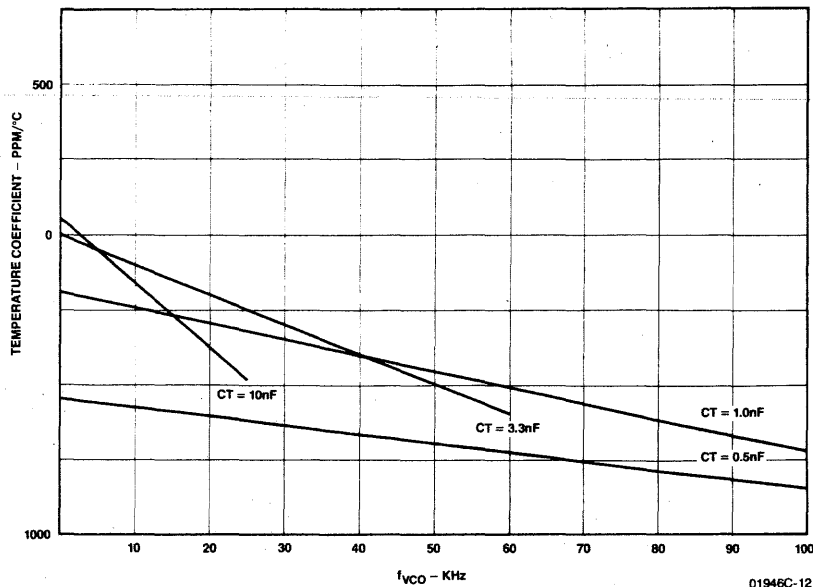
# Advanced Micro Devices

## LINEAR Am6301

### FUNCTIONAL PIN DESCRIPTION

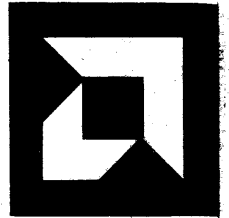
<b>V<sub>REF</sub></b>	The output of the internal 2.5V reference.		
<b>Output Q<sub>1</sub>, Q<sub>2</sub></b>	The open collector output transistors capable of sinking 70mA each.		feed-forward control works by changing the voltage across R <sub>ramp</sub> which affects the ramp slope and causes a modulation of the duty cycle.
<b>SYMQ<sub>1</sub>, Q<sub>2</sub></b>	Transformer unbalance control inputs. With a small amount of external circuitry these inputs can be used to cause asymmetrical output duty cycles. This allows correction for imbalances in the push-pull circuitry.	<b>Comp In</b>	The noninverting input of the pulse width modulating comparator.
<b>SYNC OUT</b>	The VCO output, used for synchronizing other circuits to the Am6301.	<b>Amp Out</b>	The output of the uncommitted operational amplifier. This amplifier is used as the error amplifier in most systems.
<b>C<sub>soft start</sub></b>	An external capacitor at this pin causes the output duty cycle to increase linearly during power up.	<b>(+) Amp In, (-) Amp In</b>	These are the inputs of the uncommitted operational amplifier.
<b>R<sub>T</sub>, C<sub>T</sub></b>	An external resistor and a capacitor at these pins control the VCO center frequency of the phase locked loop oscillator.	<b>ON/OFF Under Voltage</b>	This input disables the outputs when it is connected to a voltage lower than V <sub>REF</sub> . It can be connected as a remote shutdown or as an under voltage protection.
<b>C<sub>filter</sub></b>	An external filter capacitor for the control voltage of the phase locked loop is connected to this pin.	<b>Over Voltage Input</b>	This input disables the outputs whenever it is higher than V <sub>REF</sub> .
<b>SYNC IN</b>	This pin is connected to one input of the phase comparator in the phase locked loop, the other input is internally connected to the VCO output.	<b>Over Voltage Output</b>	The output of the over voltage comparator. This pin can be connected back to the over voltage input for SCR-type protection.
<b>R<sub>R</sub>, C<sub>R</sub></b>	An external resistor and a capacitor at these pins control the slope of the ramp generator. The	<b>+I<sub>DYN</sub>, -I<sub>DYN</sub></b>	Sense inputs for the dynamic current limiting circuit.

Figure 5. Am6301 f<sub>VCO</sub> Temperature Coefficient.



01946C-12

# Advanced Micro Devices



## COMMUNICATIONS Am7910 PRELIMINARY DATA (Revised)

### Am7910 FSK Modem WORLD-CHIP™

- Complete FSK MODEM in a 28-pin package (except line interface)
- Meets basic Bell 103/113/108, Bell 202, CCITT V.21, CCITT V.23 specifications (pin-programmable selection)
- No external filtering required
- All digital processing, including digital filters
- ADC/DAC on chip
- Includes essential RS-232/CCITT V.24 handshake signals
- Auto-answer capability
- Local copy/test modes
- 1200 bps full duplex on 4-wire line

### GENERAL DESCRIPTION

The Am7910 is a single-chip asynchronous Frequency Shift Keying (FSK) voiceband modem. Operating at rates up to 300, 600 or 1200 bits per second, it is compatible with the applicable Bell and CCITT recommended standards for 103/113/108, 202, V.21 and V.23 type modems. Five mode control lines select a desired modem configuration.

Digital signal processing techniques are employed in the Am7910 to perform all major functions such as modulation, demodulation and filtering. The Am7910 contains on-chip analog-to-digital and digital-to-analog converter circuits to minimize the external components in a system. This device includes the essential RS-232/CCITT V.24 terminal control signals with TTL levels.

Clocking can be generated by attaching a crystal to drive the internal crystal oscillator or by applying an external clock signal.

A data access arrangement (DAA) or acoustic coupler must provide the phone line interface externally.

The Am7910 is fabricated using N-channel MOS technology in a 28-pin package. All the digital input and output signals (except the external clock signal) are TTL compatible. Power supply requirements are  $\pm 5$  volts.

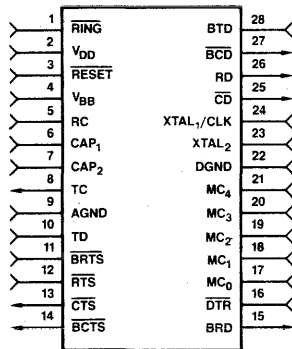


Figure 1. Am7910 Pinout

MMC-036

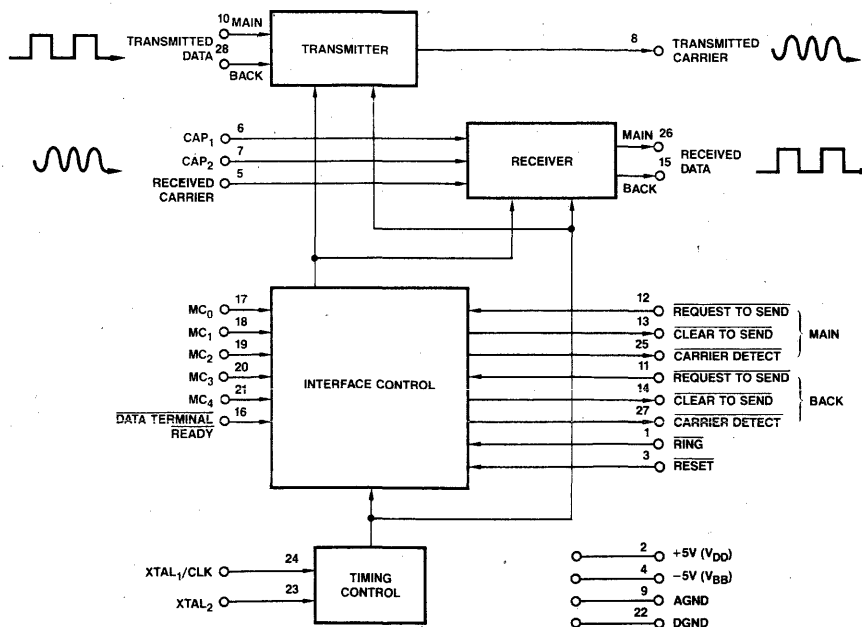
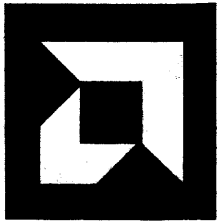


Figure 2. Am7910 Block Diagram

MMC-179

WORLD-CHIP is a trademark of Advanced Micro Devices.





# Advanced Micro Devices

## COMMUNICATIONS Am7910

### A COMPLETE FSK MODEM ON ONE WORLD-CHIP™

#### It's the World's First

The new Am7910 is the first complete, asynchronous Frequency Shift Keying modem ever offered on a single LSI chip. Our Am7910, a crystal, and a few inexpensive, non-critical components are all you need. No external filters, no hybrids, no tuned circuits are needed. Never before has the modem function been so easy to build into your products. See Figure 1.

#### It's the World's Most Complete

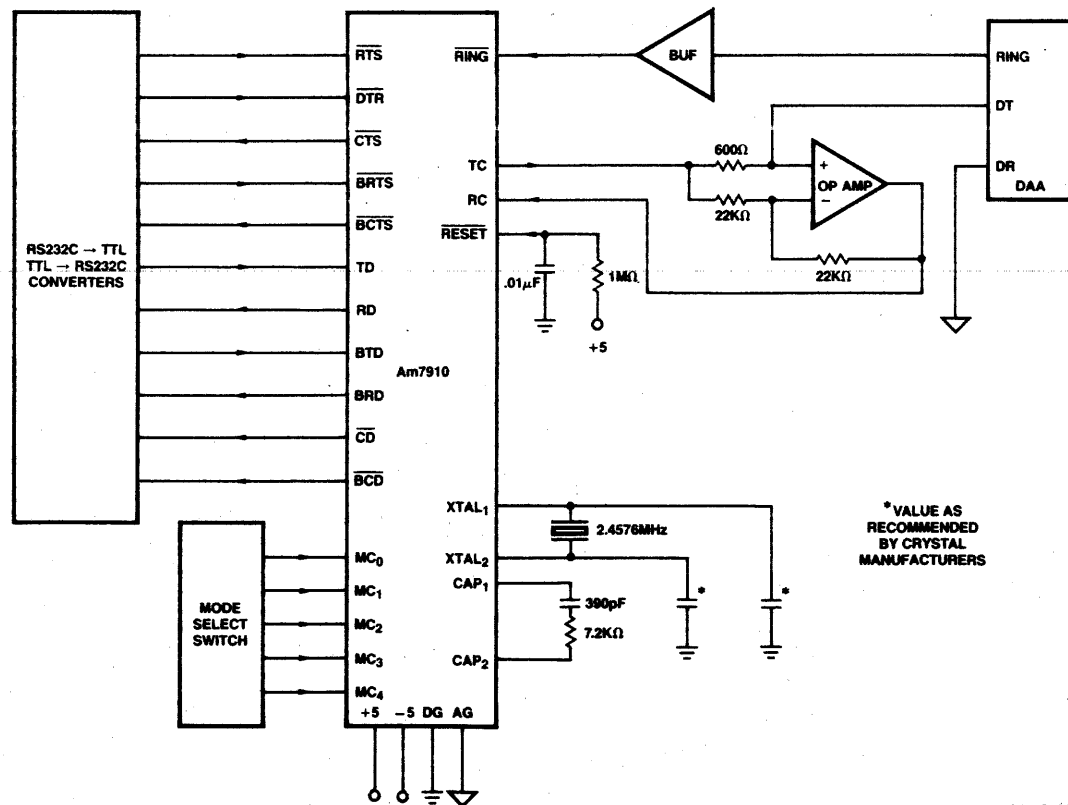
All the features a modem should have are built right in. Filters are already there. Handshake signals already there. Auto-answer is already there. Local loopback is already there. Back channels are already

there. No extra components are required to implement these functions. Even line equalization is available if you want it. No modem chip or chip set has ever offered you so many features.

#### It's a WORLD-CHIP™

No matter where you market your product, the Am7910 is the perfect modem solution. It's designed to meet communications performance standards around the world. Without any additional circuitry, it can be switched to any of 9 Bell or CCITT standards. The flexibility of Digital Signal Processing allows systems built with the Am7910 to be used all over the world without modification to the modem circuit.

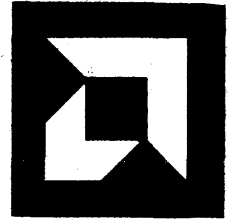
Figure 1. Stand-Alone Am7910 Application



MMC-224

LINEAR

Advanced Micro Devices



## DIGITAL SIGNAL PROCESSING FOR HIGH STABILITY, WITHOUT CRITICAL COMPONENTS

### The Am7910 Doesn't Drift With Time or Temperature

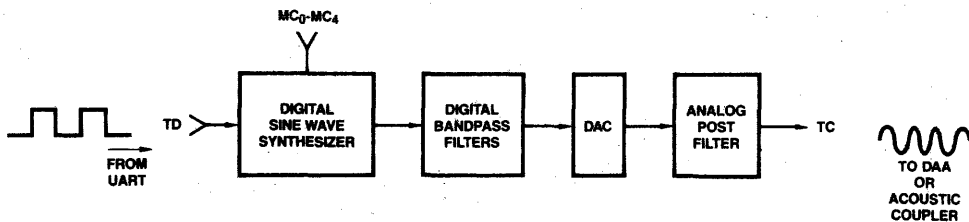
The Am7910 uses digital signal processing techniques (DSP) to perform all major functions, including modulation, demodulation and filtering. Because of the use of DSP, the performance of each section of the Am7910 is perfectly predictable and inherently stable over time and temperature. The modem performance specifications are designed into the digital signal processing algorithms, and that's where they stay, inside the chip, never changing, guaranteed. There's nothing to adjust, nothing to age or drift. See Figure 2.

### No External Filters

Digital Signal Processing completely eliminates the need for external critical-tolerance filters that analog designs require. Since no critical tuned components are required, printed circuit board design with the Am7910 is easier than ever. And DSP does not produce the noise and supply voltage sensitivity that is commonly found in switched capacitor designs. The bottom line is, fewer components, therefore a lower cost system solution.

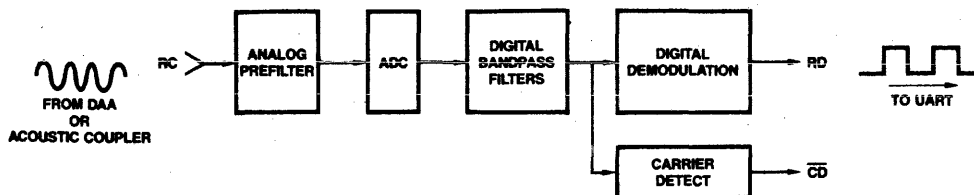
Figure 2.

#### a) Transmitter Block Diagram

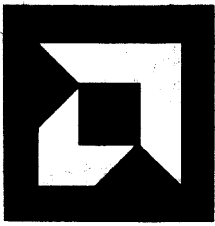


MMC-033

#### b) Receiver Block Diagram



MMC-034



# Advanced Micro Devices

## COMMUNICATIONS Am7910

### SELECTABLE TO ANY OF 9 DIFFERENT WORLD STANDARDS

The Am7910 is designed to be used in equipment all over the world. The device operates in Bell 103 and 202, and CCITT V.21 and V.23 modem configurations at baud rates from 300 to 1200bps

(with back channel). Mode selection is controlled by five simple programming inputs. No crystals need to be changed. No extra resistors, capacitors, or interface circuits. Just switch a few control lines.

### MODEM CONFIGURATIONS

STD	BPS	DUPLEX	FEATURES
Bell 103	300	Full	Originate
Bell 103	300	Full	Answer
Bell 202	1200	Half	
Bell 202	1200	Half	Line Equalizer
CCITT V.21	300	Full	Originate
CCITT V.21	300	Full	Answer
CCITT V.23 mode 2	1200	Half	
CCITT V.23 mode 2	1200	Half	Line Equalizer
CCITT V.23 mode 1	600	Half	

### MORE BUILT-IN FEATURES MAKE THE Am7910 THE EASIEST MODEM TO USE

#### Loopback for Simplified Testing

Ten loopback modes exist which permit both analog and digital loopback for each modem configuration. When a loopback mode is selected, the signal processing circuits for both the transmitter and receiver are set to operate on the same channel or frequency band. The analog output (Transmitted Carrier) and the analog input, (Received Carrier) can be externally connected together for local analog loopback. Alternatively, the digital data signals (TD and RD, or BTD and BRD,) can be connected externally, allowing a remote modem to test the local modem with its digital data signals looped back. Thus, the Am7910 reduces maintenance, service time, and cost. The 202 and V.23 loopback modes can also be used for 1200bps full duplex operation over 4-wire lines.

#### Back Channel Option Included

Low bit-rate back channels are provided by the Am7910 when the Bell 202 or CCITT V.23 modem configurations are selected. The back channel uses the remaining bandwidth of the line to return acknowledgement and control signals to the sender on another channel, while the sender continues to transmit at 1200 baud. Overall transmission speed is improved by not having to turn the line around to send

an acknowledgement, nor is it necessary to complete transmission of a data block before receiving an error signal. The 202 back channel allows up to 5 bits per second and the V.23 back channel allows up to 75 bits per second.

#### Auto-Answer for Remote Installations

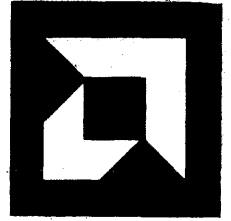
Auto-answer capability, important for equipment in remote places, is also built in. Upon receipt of a signal at its Ring Input, the Am7910 generates a silence interval followed by an answer tone of the proper duration at the Transmitted Carrier output. The auto-answer sequence meets Bell and V.25 specifications.

#### RS-232 and CCITT V.24 Terminal Control Signals Provided On-Chip

Essential terminal control signals such as Data Terminal Ready (DTR), Request to Send (RTS), Clear to Send (CTS) and Carrier Detect (CD), are included in the Am7910. All the specified delays, such as Request-to-Send ON to Clear-to-Send ON are automatically inserted. The control signals are TTL compatible.

LINEAR

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## COMMUNICATIONS Am7910

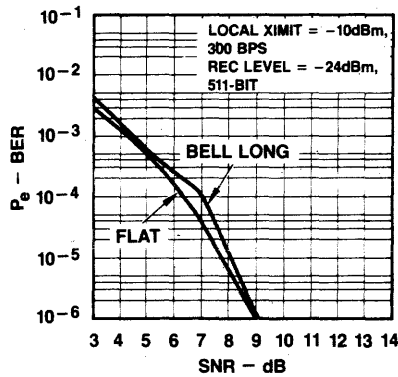
### SETS THE INTERNATIONAL STANDARD OF PERFORMANCE

The measure of a modem's performance is its ability to extract correct data from a signal received over a line with severe distortion, attenuation, and noise. The graphs below show the Bit Error Rate (BER) of the Am7910 under two sets of conditions. At the left is the data for a Bell 103 configuration over an

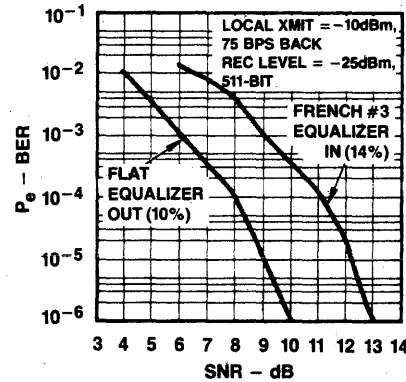
undistorted line and a line with amplitude and group delay distortion. On the right is the data for a V.23 configuration for an undistorted and a severely distorted line. Compare this performance data with any other modem chip or chip set and you'll see why the Am7910 sets the standard.

### Performance Curves

**Figure 3. 103 ORIG, 300bps BER versus SNR, Different Lines**



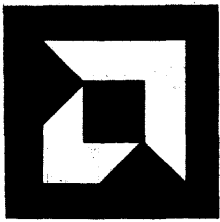
**Figure 4. CCITT V.23, 1900bps BER versus SNR, Different Lines**



MMC-054

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# Advanced Micro Devices

## COMMUNICATIONS Am7910

### THE WORLD-CHIP™ SERIES

Telephone systems around the world are evolving rapidly into something new: a huge digital data communications network. The crossbar switches of yesterday are being replaced by compact electronic switches that route digital bit streams from one subscriber to another. The wires that used to carry human speech alone now also carry streams of digital data via modems. Soon the telephone set itself will be changed from a simple device for talking and listening into a specialized data terminal, capable of doing much more.

As the information carried by our telephone networks has changed, new standards have been added to the old. Standards to digitize voice. Standards for sending digital data over voice-grade lines. Standards for multiplexing many data packets onto a single line. Around the world, two groups of standards are used — one developed by Bell in the U.S., another developed by the CCITT. Until now, the electronic components designed to make this revolution possible have been specialized for one group of standards or the other, and usually for only one standard or variation within that group.

Advanced Micro Devices' World-Chip™ Series is different. We're using a variety of technologies to create circuits that support the telecom revolution all over the world. We're using advanced architectural

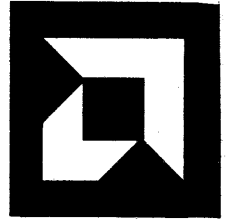
concepts, especially digital signal processing, to move critical filtering and signal conditioning directly onto the integrated circuit. Digital techniques provide guaranteed high stability and repeatability, and also allow changes to filtering characteristics under user program control. We're using our skills in both linear and digital circuit design to put the conversions between analog signals and digital data on the same IC as the digital processing circuits. And we're using our MOS and Bipolar process technologies to solve each part of the system problem in an optimum way with parts that are designed to work together. By the end of the decade AMD will have led the way to more efficient, effective and inexpensive methods for making the world an even smaller (and faster) place in which to communicate.

AMD has developed a set of components that work in any of the world's telecommunications systems. Each integrated circuit can be programmed to perform its function in conformance with any of the applicable standards. The state-of-the-art features built into our devices are available simultaneously to customers designing equipment for any of the world's telecom markets. And customers who themselves design equipment for more than one market can use a single set of circuits — Advanced Micro Devices' World-Chips™ — to meet all their needs.

### RELATED WORLD-CHIP™ SERIES PRODUCTS:

- Am7901/7902 Subscriber Line Audio-Processing Circuit
- Am7950 SLIC

# Advanced Micro Devices



## COMMUNICATIONS Am7990 Family Ethernet Node

Advanced Micro Devices has developed a set of MOS and Bipolar high performance integrated circuits designed to minimize the cost of connecting multivendor devices to an Ethernet bus. This family is designed to provide in the minimum package count, all the logic, protocol and control functions required to interface all the popular 16-bit processors to an Ethernet local area network.

### **Am7990 Local Area Network Controller for Ethernet (LANCE)**

- Buffer management
- On board DMA
- Limit Error Detection
- Address Detection
- Line Access Protocol (CSMA/CD)
- Collision handling

### **Am7991 Serial Interface Adaptor (SIA)**

- Manchester encoding/decoding
- Differential to TTL signal conversion
- Transceiver cable interface

### **Am79XX Ethernet Transceiver**

- Collision detection
- Line protection

## **Key System Level Features**

### **16-Bit Bus Interface**

#### **Compatible with**

- Z8000
- 8086
- 68000
- LSI-II

### **Broad Range of Diagnostics**

### **Alternate Sourced**

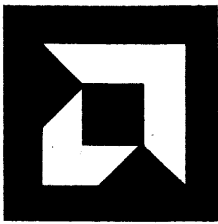
### **Meets the Ethernet Specification**

#### **Ethernet Data Link Layer Support**

- Buffer management
- Data encapsulation
- Framing and packet control
- CRC generation/check
- Serial/deserialization

#### **Ethernet Physical Link Layer Support**

- Data rate: 10 Mb/s
- Carrier-sense multiple-access with collision detection
- Transceiver interface compatibility



# Advanced Micro Devices

## COMMUNICATIONS Am7990 Family Ethernet Node System Overview

Advanced Micro Devices is introducing a set of LSI devices that provides the Ethernet system designer, and designers of Ethernet compatible products, a low cost physical and link level interface to the Ethernet Bus.

The Am7990 Ethernet interface family is being designed using a combination of MOS and IMOX™ bipolar technologies. This family consists of the Am7990 Local Area Network Controller for Ethernet (LANCE), and the Am7991 Serial Interface Adapter (SIA). As shown in Figure 1, the Am7990 family provides the complete interface between the device System Bus and the Ethernet Transceiver Cable.

The Am7990 LANCE is a 10M-bit/sec MOS device in a 48-pin package, optimized to perform the link level Ethernet protocol. The CSMA/CD network access, memory management (onboard DMA), error reporting, packet handling, and microprocessor interface functions also reside in the LANCE.

The Am7991 Serial Interface Adaptor provides Manchester encoding and decoding of the serial bit stream and interfaces the TTL output of the LANCE to the differential inputs of the transceiver. It has an on board phase locked loop to recover clock from an incoming signal and can use an external crystal oscillator or TTL inputs to provide clock for transmission.

Coupling the Ethernet Node to the Ethernet Cable requires a transceiver. Commercially available board or module transceivers can be used with the LANCE and SIA. Advanced Micro Devices has a monolithic transceiver in an early phase of development which should lower the cost significantly in this area as well.

### BASIC SYSTEM OPERATION

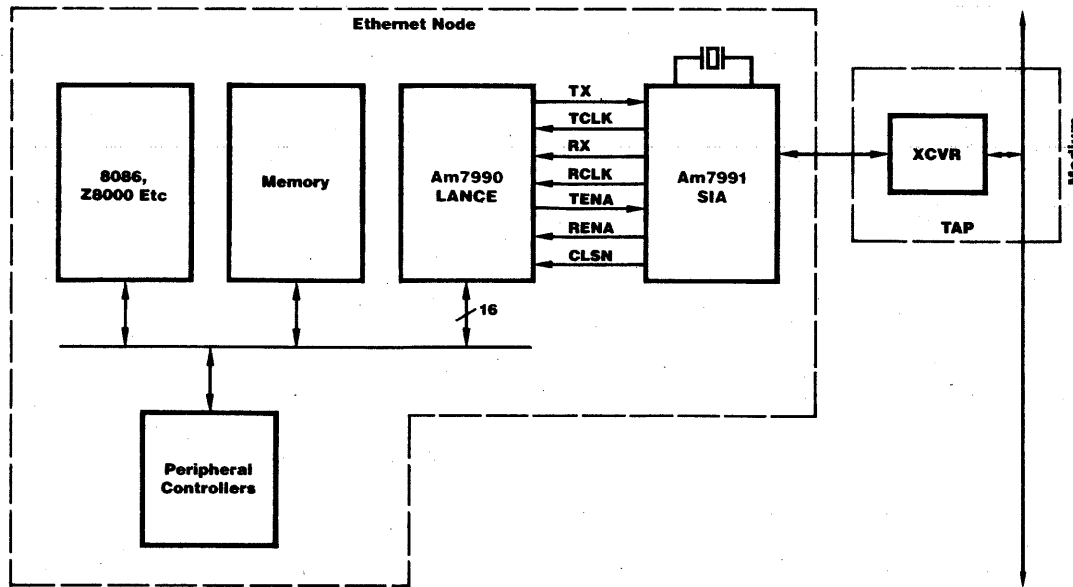
Ethernet is a send and receive half duplex system. The node must function in either transmit or receive mode at any instant in time. Before transmission the node must be sure there is no contention for the bus. The Ethernet CSMA/CD network access algorithm is implemented completely within the LANCE. In addition to listening for a clear coax before transmitting, Ethernet handles collisions in a predetermined way. Should two nodes attempt to transmit at the same time, the signals will collide and the data on the coax will be garbled. The transmitting nodes listen while they transmit and detect the collision. Both continue to transmit for a predetermined length of time to "jam" the network, insuring all nodes have recognized the collision. The transmitting nodes then delay a random amount of time according to the "truncated binary backoff" algorithm implemented in the LANCE, before attempting to transmit again. This minimizes the possibility of collision on retransmission.

### TRANSMIT MODE

In the transmit mode, the LANCE initiates a DMA cycle to access data from a transmit buffer. It prefaces the data with a preamble, and sync pattern then calculates and appends a 32-bit CRC.

This packet is transmitted serially to the SIA. The Manchester encoder in the SIA takes the transmitted data from the LANCE and creates the Manchester encoded differential signals TRANSMIT+ and TRANSMIT- to drive the Transceiver cable. These differential signals are coupled through the transceiver cable, transceiver and on to the Ethernet coaxial cable.

Figure 1. Ethernet Node Architecture

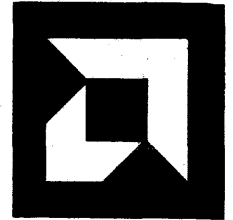


MMC-226

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## COMMUNICATIONS Am7990 Family Ethernet Node

### RECEIVE MODE

When carrier is present on the Ethernet coax, the Transceiver will create the differential signals RECEIVE+ and RECEIVE-. These inputs to the SIA are decoded by the Manchester decoder. A phase locked loop synchronizes to the Ethernet Preamble, allowing the decoder to recover clock and data from the encoded signals. These two signals are supplied to the LANCE as the TTL signals RECEIVE DATA and RECEIVE CLOCK. In addition, the SIA creates the signal CARRIER PRESENT while it is receiving data from the cable, indicating to the LANCE that receive data and clock are available. When these signals reach the LANCE, the CRC is calculated and compared to the CRC checksum at the end of the packet. If the calculated CRC doesn't agree with the packet CRC an error bit is set and an interrupt generated to the microprocessor.

### ADDRESSING

There are three addressing modes. The first is physical addressing which requires a comparison of the 48-bit destination address in the packet with the node address programmed into the LANCE during initialization. The second mode is multi-cast addressing. This mode can be useful when sending packets to all of one type of a device simultaneously on the network, or for a broadcast

situation where all nodes on the network receive the packet. In the final "promiscuous" mode of operation, a node will accept all packets on the coax regardless of their destination address.

### ERROR REPORTING

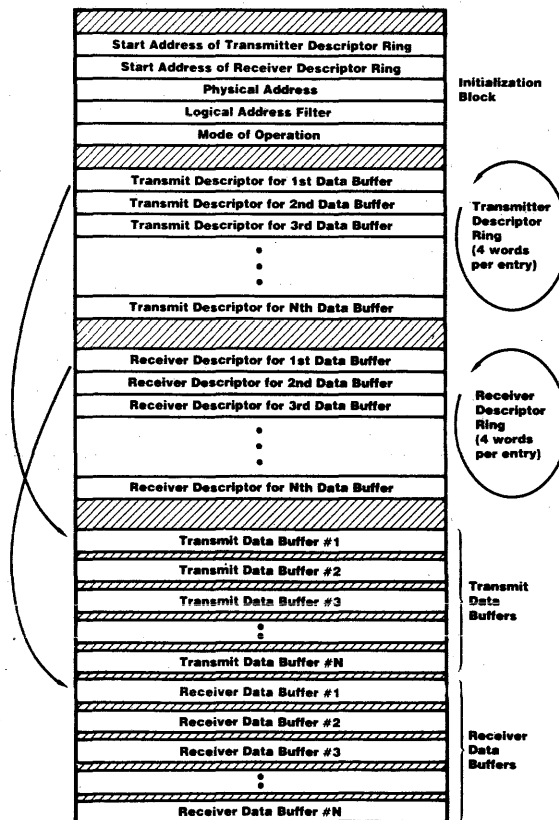
Extensive error reporting is provided by the LANCE through microprocessor interrupt and error bits in a status register. The following are the significant error conditions:

- CRC error on receive
- Babbling error
- Missed packet
- Memory error

### BUFFER MANAGEMENT

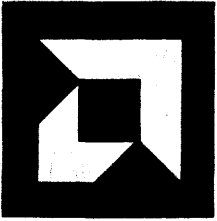
A key feature of the LANCE and it's on board DMA channel is the flexibility and speed of communication between the LANCE and the host microprocessor through common memory locations. The basic organization of the buffer management consists of circular task queues called descriptor rings for transmit and receive operations. Up to 128 tasks may be queued on a descriptor ring awaiting execution by the LANCE. (Figure 2)

Figure 2. LANCE/Processor Memory Interface



MMC-105





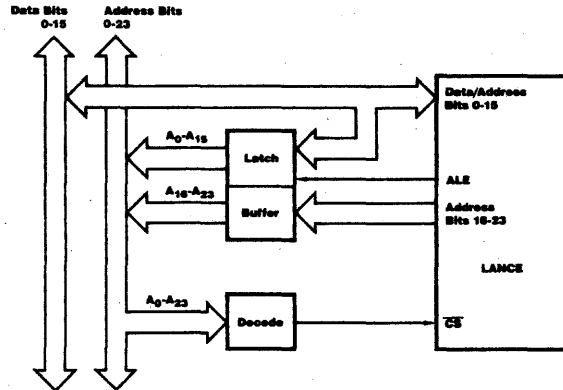
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## COMMUNICATIONS Am7990 Family Ethernet Node

### MICROPROCESSOR INTERFACE

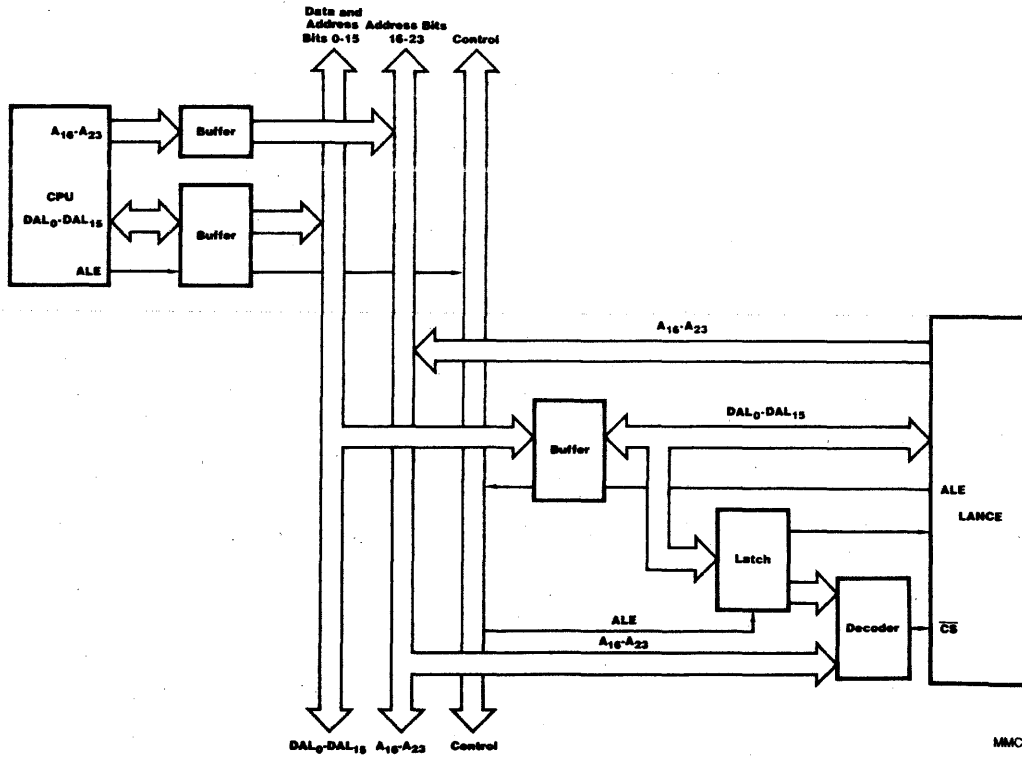
The parallel interface of the LANCE has been designed to be "friendly" or easy to interface to a variety of popular 16-bit microprocessors. These microprocessors include the 68000, Z8000, 8086, and LSI-II devices.

The LANCE has a 24-bit wide linear address space when it is in the Bus Master Mode allowing it to DMA directly into the entire address space of the above microprocessors. The LANCE interfaces with both multiplexed and demultiplexed data busses (Figure 3) and features control signals for address/data bus transceivers.



MMC-106

a) Demultiplexed Bus



MMC-107

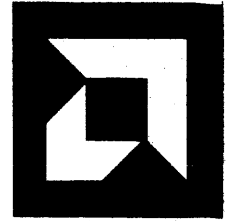
b) Multiplexed Bus System Overview

Figure 3.

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## COMMUNICATIONS Am7990 Family Ethernet Node IN DEVELOPMENT

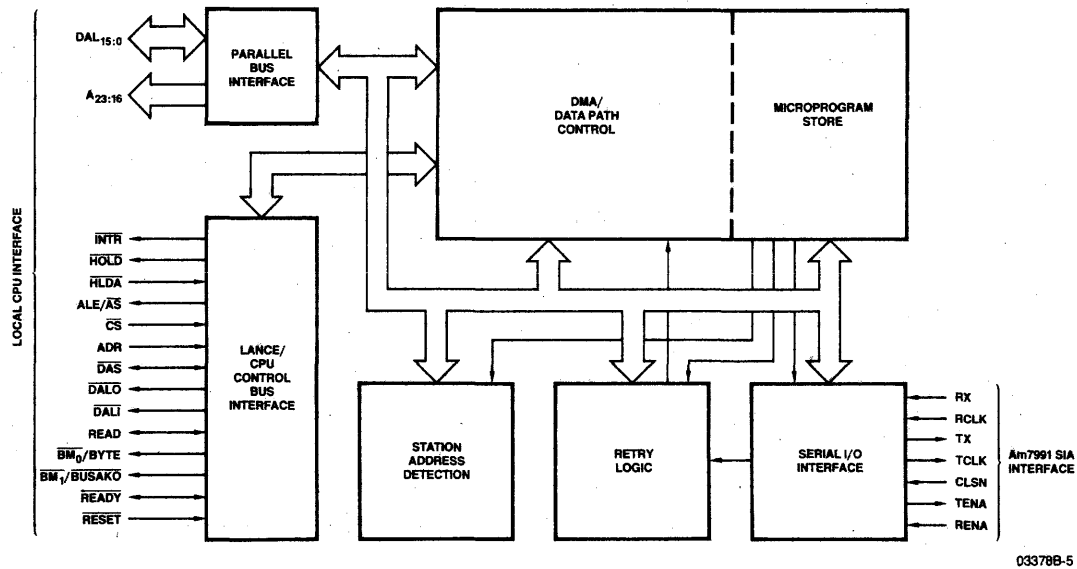
### DISTINCTIVE CHARACTERISTICS

- IEEE 802.3 compatible
- Easily interfaced to 8086, 68000, Z8000, LSI-II microprocessors
- On-board DMA and buffer management
- 24-bit wide linear addressing (Bus Master Mode)
- Network and packet error reporting
- Diagnostic Routines
  - Internal/external loop back
  - CRC logic check
  - Time domain reflectometer

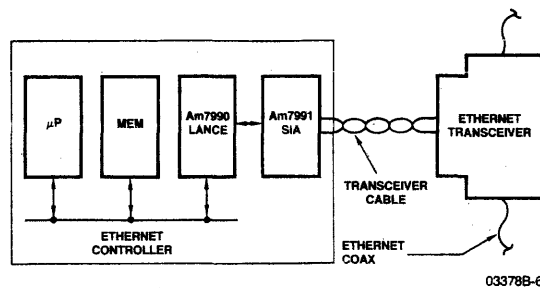
### GENERAL DESCRIPTION

The Am7990 Local Area Network Controller for Ethernet (LANCE) is a 48-pin VLSI device designed to greatly simplify interfacing a microcomputer or minicomputer to an Ethernet Local Area Network. This chip, in conjunction with the Am7991 Serial Interface Adapter (SIA) and closely coupled local memory and microprocessor, is intended to provide the user with a complete interface module for an Ethernet network. The Am7990 is designed using a scaled N-Channel MOS technology and is compatible with a variety of microprocessors. On-board DMA, advanced buffer management and extensive error reporting and diagnostics facilitate design and improve system performance.

### LANCE BLOCK DIAGRAM

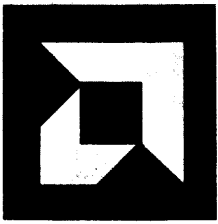


### TYPICAL ETHERNET NODE



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## COMMUNICATIONS Am7990 Family Ethernet Node

### PIN DEFINITIONS

**DAL<sub>00</sub>–DAL<sub>15</sub>** **Data/Address Lines (Input/Output 3-State).** The time multiplexed Address/Data bus. During the address portion of a memory transfer, DAL<sub>00</sub>–DAL<sub>15</sub> contains the lower 16 bits of the memory address. The upper 8 bits of address are contained in A<sub>16</sub>–A<sub>23</sub>.

During the data portion of a memory transfer, DAL<sub>00</sub>–DAL<sub>15</sub> contains the read or write data, depending on the type of transfer.

The LANCE drives these lines as a Bus Master and as a Bus Slave.

**A<sub>16</sub>–A<sub>23</sub>** **High Order Address Bus (Output 3-State).** The additional address bits necessary to extend the DAL lines to access a 24-bit address. These lines are driven as a Bus Master only.

**READ** **(Input/Output 3-State).** Indicates the type of operation to be performed in the current bus cycle. This signal is an output when the LANCE is a Bus Master.

- High – Data is taken off the DAL by the chip
- Low – Data is placed on the DAL by the chip

The signal is an input when the LANCE is a Bus Slave

- High – Data is placed on the DAL by the chip
- Low – Data is taken off the DAL by the chip

**$\overline{BM}_0$ /BYTE** **I/O pins 15 and 16 are programmable through bit (00) of CSR<sub>3</sub>.**

**$\overline{BM}_1$ /BUSAKO** **If CSR<sub>3</sub> (00) BCON = 0**  
I/O PIN 15 =  $\overline{BM}_0$  (Output 3-state)  
I/O PIN 16 =  $\overline{BM}_1$  (Output 3-state)

$\overline{BM}_0$ ,  $\overline{BM}_1$  (Byte Mask). This indicates the byte(s) on the DAL are to be read or written during this bus transaction. The LANCE drives these lines only as a Bus Master. It ignores the Byte Mask lines when it is a Bus Slave, and assumes word transfers.

Byte selection using Byte Mask is done as described by the following table.

$\overline{BM}_1$	$\overline{BM}_0$	
LOW	LOW	Whole Word
LOW	HIGH	Upper Byte
HIGH	LOW	Lower Byte
HIGH	HIGH	None

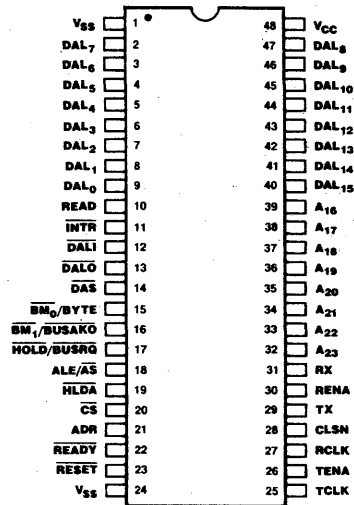
**If CSR<sub>3</sub> (00) BCON = 1**  
I/O PIN 15 = BYTE (Output 3-state)  
I/O PIN 16 = BUSAKO (Output)

Byte selection may also be done using the BYTE line and DAL<sub>00</sub> line, latched during the address portion of the bus cycle. The LANCE drives BYTE only as a Bus Master and ignores it when a Bus Slave selection is done (similar to  $\overline{BM}_0$ ,  $\overline{BM}_1$ ).

Byte selection is done as outlined in the following table.

BYTE	DAL <sub>00</sub>	
LOW	LOW	Whole Word
LOW	HIGH	Illegal Condition
HIGH	LOW	Lower Byte
HIGH	HIGH	Upper Byte

### LANCE PIN DESCRIPTION D-48-1



Note: Pin 1 is marked for orientation.

**BUSAKO** is a bus request daisy chain output. If the chip is not requesting the bus and it receives HLDA, BUSAKO will be driven low. If the LANCE is requesting the bus when it receives HLDA, BUSAKO will remain high.

### Byte Swapping

In an effort to be compatible with the variety of 16-bit microprocessors available to the designer, the LANCE may be programmed to swap the position of the upper and lower order bytes on data involved in transfers with the internal FIFO.

Byte swapping is done when BSWP = 1. The most significant byte of the word in this case will appear on DAL lines 7-0 and the least significant byte on DAL lines 15-8.

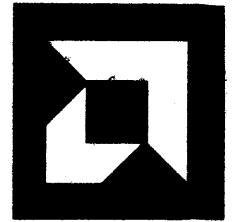
When BYTE = H (indicating a byte transfer) the table indicates on which part of the 16 bit data bus the actual data will appear.

Whenever byte swap is activated, the only data that is swapped is data traveling to and from the FIFO.

Signal Line	Mode Bits	BSWP = 0 and BCON = 1	BSWP = 1 and BCON = 1
	BYTE = L and DAL <sub>00</sub> = L		Word
BYTE = L and DAL <sub>00</sub> = H		Illegal	Illegal
BYTE = H and DAL <sub>00</sub> = H		Upper Byte	Lower Byte
BYTE = H and DAL <sub>00</sub> = L		Lower Byte	Upper Byte

**$\overline{CS}$**  **Chip Select (Input).** Indicates, when asserted, that the LANCE is the slave device of the data transfer.  $\overline{CS}$  must be valid throughout the data portion of the bus cycle.

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## COMMUNICATIONS Am7990 Family Ethernet Node

<b>ADR</b>	<b>Register Address Port Select (Input).</b> When LANCE is slave, ADR indicates which of the two register ports is selected. ADR LOW selects register data port, ADR HIGH selects register address port. ADR must be valid throughout the data portion of the bus cycle and is only used by the LANCE when CS is low.		
<b>ALE/<math>\overline{AS}</math></b>	<b>Address Latch Enable (Output 3-State).</b> Used to demultiplex the DAL lines and define the address portion of the bus cycle. This I/O pin is programmable through bit (01) of CSR <sub>3</sub> .  As ALE, (CSR <sub>3</sub> (01), ACON = 0) the signal transitions from a HIGH to a LOW during the address portion of the transfer and remains low during the data portion. ALE can be used by a Slave device to control a latch on the bus address lines. When ALE is high the latch is open and when ALE goes low the latch is closed.  As $\overline{AS}$ (CSR <sub>3</sub> (01), ACON = 1), the signal pulses LOW during the address portion of the bus transaction. The low to high transition of AS can be used by a slave device to strobe the address into a register.  The LANCE drives the ALE/ $\overline{AS}$ line only as a Bus Master.	<b>HLDA</b>	When CSR <sub>3</sub> (00) BCON = 0 I/O pin 17 = $\overline{HOLD}$ (Output Open Drain)  When CSR <sub>3</sub> (00) BCON = 1 I/O pin 17 = $\overline{BUSRQ}$ (Output Open Drain)  $\overline{BUSRQ}$ will be asserted only if I/O pin 17 is high prior to assertion.
<b><math>\overline{DAS}</math></b>	<b>Data Strobe (Input/Output 3-State).</b> Defines the data portion of the bus transaction. $\overline{DAS}$ is high during the address portion of a bus transaction and low during the data portion. The low to high transition can be used by a Slave device to strobe bus data into a register. $\overline{DAS}$ is driven only as a Bus Master.	<b>INTR</b>	<b>Bus Hold Acknowledge (Input).</b> A response to $\overline{HOLD}$ . When HLDA is low in response to the chip's assertion of $\overline{HOLD}$ , the chip is the Bus Master. HLDA deasserts upon the deassertion of $\overline{HOLD}$ .  <b>Interrupt (Output Open Drain).</b> An attention signal that indicates, when active, that one or more of the following CSR <sub>0</sub> status flags is set: BABL, MERR, MISS, RINT, TINT or IDON. INTR is enabled by bit 06 of CSR <sub>0</sub> (INEA = 1).
<b>DALO</b>	<b>Data/Address Line Out (Output 3-State).</b> An external bus transceiver control line. DALO is asserted when the LANCE drives the DAL lines. DALO will be low only during the address portion if the transfer is a READ. It will be low for the entire transfer if the transfer is a WRITE. DALO is driven only when LANCE is a Bus Master.	<b>RX</b>	<b>Receive (Input).</b> Receive Input Bit Stream.
<b>DALI</b>	<b>Data/Address Line In (Output 3-State).</b> An external bus transceiver control line. DALI is asserted when the LANCE reads from the DAL lines. It will be low during the data portion of a READ transfer and remain high for the entire transfer if it is a WRITE. DALI is driven only when LANCE is a Bus Master.	<b>TX</b>	<b>Transmit (Output).</b> Transmit Output Bit Stream.
<b>HOLD/<math>\overline{BUSRQ}</math></b>	<b>Bus Hold Request (Output Open Drain).</b> Asserted by the LANCE when it requires access to memory. $\overline{HOLD}$ is held LOW for the entire ensuing bus transaction. The function of this pin is programmed through bit (00) of CSR <sub>3</sub> . Bit (00) of CSR <sub>3</sub> is cleared when $\overline{RESET}$ is asserted.	<b>TENA</b>	<b>Transmit Enable (Output).</b> Transmit Output Bit Stream enable. A level-asserted with the Transmit Output Bit Stream, TX, to enable the external transmit logic.
		<b>RCLK</b>	<b>Receive Clock (Input).</b> A 10MHz square wave synchronized to the Receive data and only active while receiving an Input Bit Stream.
		<b>CLSN</b>	<b>Collision (Input).</b> A logical input that indicates that a collision is occurring on the channel.
		<b>RENA</b>	<b>Receive Enable (Input).</b> A logical input that indicates the presence of carrier on the channel.
		<b>TCLK</b>	<b>Transmit Clock (Input).</b> 10MHz clock.
		<b>READY</b>	<b>(Input/Output Open Drain).</b> When the LANCE is a Bus Master, $\overline{READY}$ is an asynchronous acknowledgement from the bus memory that it will accept data in a WRITE cycle or that it has put data on the DAL lines in a READ cycle.  As a Bus Slave, the LANCE asserts $\overline{READY}$ when it has put data on the DAL lines during a READ cycle or is about to take data off the DAL lines during a write cycle. $\overline{READY}$ is a response to $\overline{DAS}$ and will return HIGH after $\overline{DAS}$ has gone HIGH. $\overline{READY}$ is an input when the LANCE is a Bus Master and an output when the LANCE is a Bus Slave.
		<b>RESET</b>	<b>(Input).</b> Bus Reset Signal. Causes the LANCE to cease operation and enter an idle state.
		<b>Vcc</b>	Power supply pin +5 volts $\pm$ 5%.
		<b>Vss</b>	Ground. Pins 1 and 24 should be connected together externally, as close to the chip as possible.

### FUNCTIONAL DESCRIPTION

The parallel interface of the Local Area Network Controller for Ethernet (LANCE) has been designed to be "friendly" or easy to interface to a variety of popular 16-bit microprocessors. These microprocessors include the following: Z8000, 8086, 68000 and LSI-11. The LANCE has a 24-bit wide linear address space when it is in the Bus Master Mode allowing it to DMA directly into the entire address space of the above microprocessors. A programmable mode of operation allows byte addressing in one of

two ways: A Byte/Word control signal compatible with the 8086 and Z8000, or an Upper Data Strobe and Lower Data Strobe signal compatible with microprocessors such as the 68000. A programmable polarity on the Address Strobe signal eliminates the need for external logic. The LANCE interfaces with both multiplexed and demultiplexed data busses and features control signals for address/data bus transceivers.

During initialization, the CPU loads the starting address of the initialization block into two internal control registers. The LANCE



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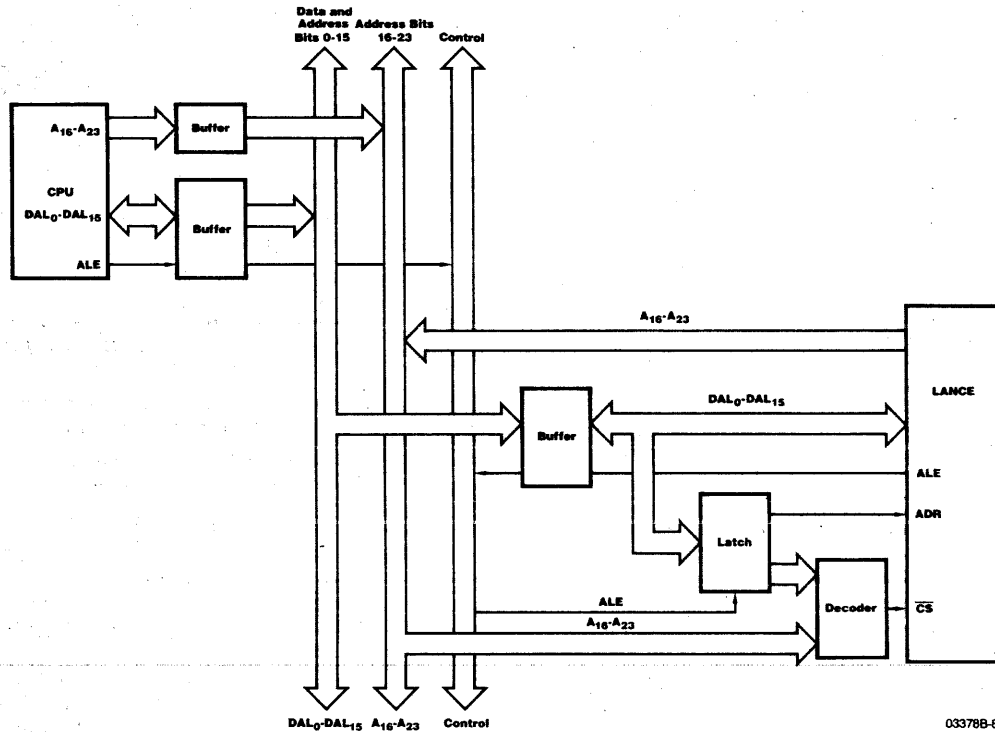
has four internal control and status registers (CSR<sub>0, 1, 2, 3</sub>) which are used for various functions such as the loading of the initialization block address, different programming modes and status conditions. The host processor communicates with the LANCE during the initialization phase, for demand transmission and periodically to read the status bits following interrupts. All other transfers to and from the memory are handled as DMA under microword control.

Interrupts to the microprocessor are generated by the LANCE: 1) upon completion of its initialization routine, 2) the reception of a packet, 3) the transmission of a packet, 4) transmitter timeout error, 5) a missed packet and 6) memory error.

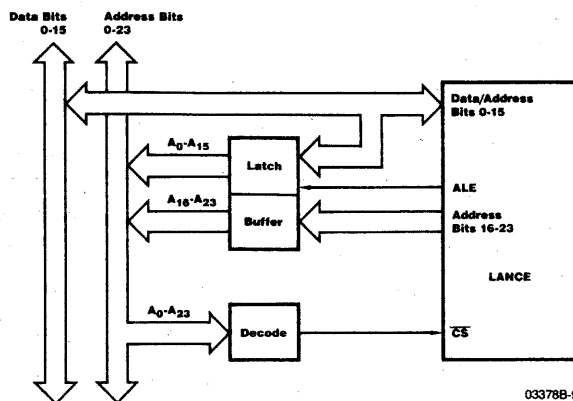
The cause of the interrupt is ascertained by reading CSR<sub>0</sub>. Bit (06) of CSR<sub>0</sub>, (INEA) enables or disables interrupts to the microprocessor. In systems where polling is used in place of

Figure 1. LANCE/CPU Interfacing

a) Multiplexed Bus

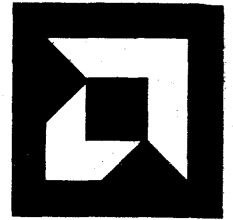


b) Demultiplexed Bus



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## COMMUNICATIONS Am7990 Family Ethernet Node

interrupts, bit (07) of CSR<sub>0</sub> (INTR) indicates an interrupt condition.

The basic operation of the LANCE consists of two distinct modes: transmit and receive. In the transmit mode the LANCE chip directly accesses data in a transmit buffer in memory. It prefaces the data with a preamble, sync pattern, and calculates and appends a 32-bit CRC. This packet is then ready for transmission to the Am7991 SIA.

In the receive mode, packets are sent via the SIA to the LANCE. The packet is loaded into buffer memory. A CRC is calculated and compared with the CRC appended to the data packet. If the calculated CRC checksum doesn't agree with the packet CRC, an error bit is set.

### ADDRESSING

Packets can be received using 3 different destination addressing schemes: physical, logical and promiscuous.

The first type is a full comparison of the 48-bit destination address in the packet with the node address that was programmed into the LANCE during an initialization cycle. There are two types of logical address, one is group type mask where the 48-bit address in the packet is put through a hash filter in order to map the 48-bit physical addresses into 1 of 63 logical groups. This mode can be useful if sending packets to all of a particular type of device simultaneously (i.e., send a packet to all file servers or all printer servers). The second logical address is a multicast address where all nodes on the network receive the packet. The last receive mode of operation is the so-called "promiscuous mode" in which a node will accept all packets on the coax regardless of their destination address.

### COLLISION DETECTION AND IMPLEMENTATION

The Ethernet CSMA/CD network access algorithm is implemented completely within the LANCE. In addition to listening for a clear coax before transmitting, Ethernet handles collisions in a predetermined way. Should two transmitters attempt to seize the coax at the same time they will collide and the data on the coax will be garbled. The transmitting nodes listen while they transmit; detect the collision, then continue to transmit for a predetermined length of time to "jam" the network and ensure that all nodes have recognized the collision. The transmitting nodes then delay a random amount of time according to the Ethernet "truncated binary backoff" algorithm in order that the colliding nodes don't try to repeatedly access the network at the same time. Up to 16 attempts to access the network are made by the LANCE before reporting back an error due to excessive collisions.

### ERROR REPORTING AND DIAGNOSTICS

Extensive error reporting is provided by the LANCE. Error conditions reported relate either to the network as a whole or to data packets. Network-related errors are recorded as flags in the CSRs and are examined by the CPU following interrupt. Packet related errors are written into descriptor entries corresponding to the packet.

System errors include:

- Babbling Transmitter
  - Transmitter attempting to transmit more than 1518 data bytes.
- Collision
  - Collision detection circuitry nonfunctional
- Missed packet
  - Insufficient buffer space
- Memory timeout
  - Memory response failure

Packet related errors:

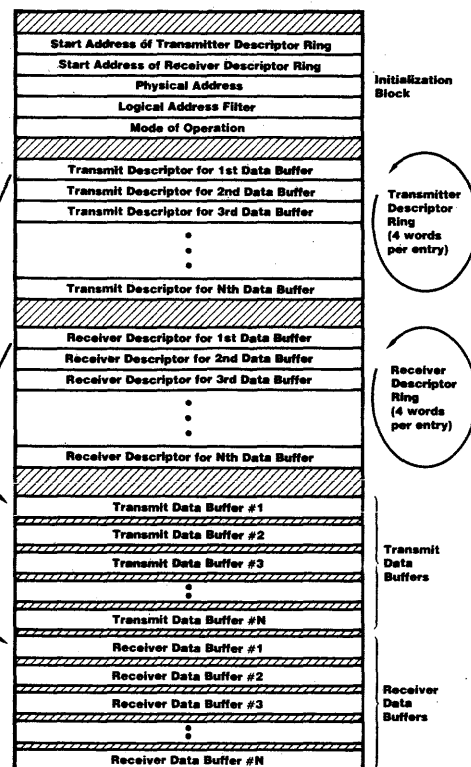
- CRC
  - Invalid data
- Framing
  - Packet did not end on a byte boundary
- Overflow/Underflow
  - Indicates abnormal latency in servicing a DMA request
- Buffer
  - Insufficient buffer space available

The LANCE performs several diagnostic routines which enhance the reliability and integrity of the system. These include a CRC logic check and two loop back modes (internal/external). Errors may be introduced into the system to check error detection logic. A Time Domain Reflectometer is incorporated into the LANCE to aid system designers locate faults in the Ethernet cable. Shorts and opens manifest themselves in reflections which are sensed by the TDR.

### BUFFER MANAGEMENT

A key feature of the LANCE and its on-board DMA channel is the flexibility and speed of communication between the LANCE and the host microprocessor through common memory locations. The basic organization of the buffer management is a circular queue of tasks in memory called descriptor rings as shown in Figure 2. There are separate descriptor rings to describe transmit and receive operations. Up to 128 tasks may be queued up on a descriptor ring awaiting execution by the LANCE. Each entry in a descriptor ring holds a pointer to a data memory buffer and an

Figure 2. LANCE/Processor Memory Interface



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## COMMUNICATIONS Am7990 Family Ethernet Node

entry for the length of the data buffer. Data buffers can be chained or cascaded in order to handle a long packet in multiple data buffer areas. The LANCE searches the descriptor rings in a "look ahead manner" to determine the next empty buffer in order to chain buffers together or to handle back to back packets. As each buffer is filled, an "own" bit is reset allowing the host processor to process the data in the buffer.

### LANCE INTERFACE

CSR bits such as ACON, BCON and BSWP are used for programming the pin functions used for different interfacing schemes. For example, ACON is used to program the polarity of the Address Strobe signal (ALE/ $\overline{AS}$ ).

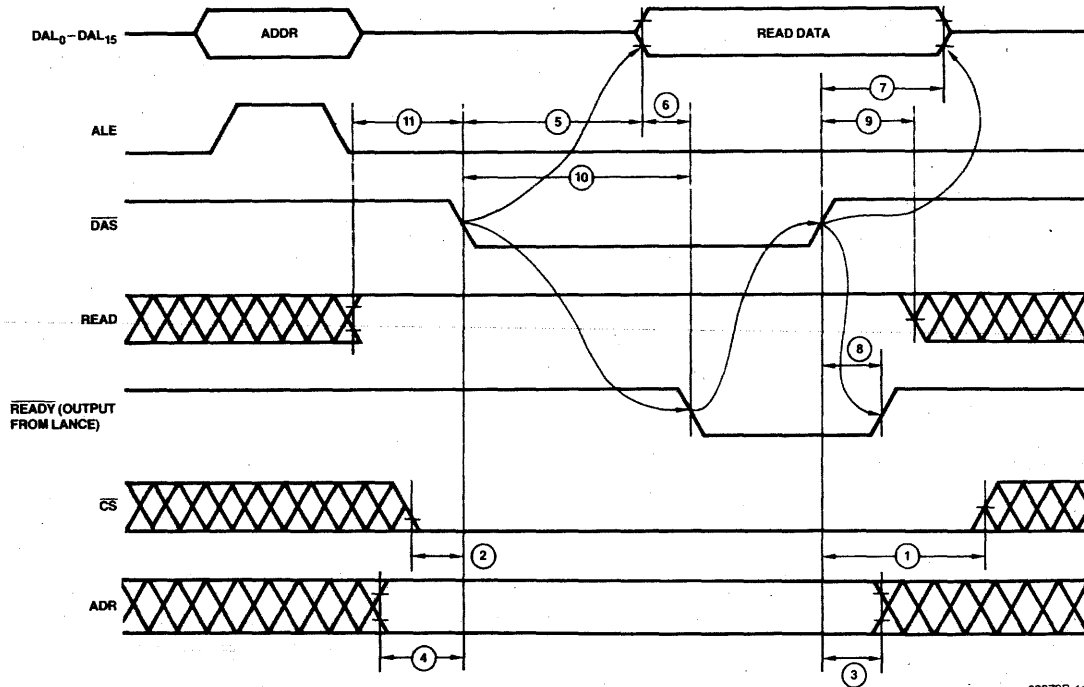
BCON is used for programming the pins for handling either the BYTE/WORD method for addressing word organized, byte addressable memories where the BYTE signal is decoded along

with the least significant address bit to determine upper or lower byte, or an explicit scheme in which two signals labeled as BYTE MASK ( $\overline{BM}_0$  and  $\overline{BM}_1$ ) indicate which byte is addressed. When the BYTE scheme is chosen the  $\overline{BM}_1$  pin can be used for performing the function  $\overline{BUSA}_{KO}$ .

BCON is also used to program pins for different DMA modes. In a daisy chain DMA scheme, 3 signals ( $\overline{BUSRQ}$ ,  $\overline{HLDA}$ ,  $\overline{BUSA}_{KO}$ ), are used. In systems using a DMA controller for arbitration, only  $\overline{HOLD}$  and  $\overline{HLDA}$  are used.

All data transfers from the LANCE in the Bus Master mode are timed by ALE,  $\overline{DAS}$  and  $\overline{READY}$ . The automatic adjustment of the LANCE cycle by the  $\overline{READY}$  signal allows synchronization with variable cycle time memory due either to memory refresh or to dual port access. Bus cycles are a minimum of 600ns in length and can be increased in 100ns increments.

### BUS SLAVE WRITE TIMING

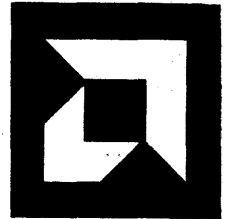


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Note: 1. There are two types of delays which depend on which internal register is accessed.  
Typ 1 refers to access of CSR<sub>0</sub>, CSR<sub>1</sub> and RAP.  
Typ 2 refers to access of CSR<sub>1</sub> and CSR<sub>2</sub>.

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## COMMUNICATIONS Am7990 Family Ethernet Node

### READ SEQUENCE

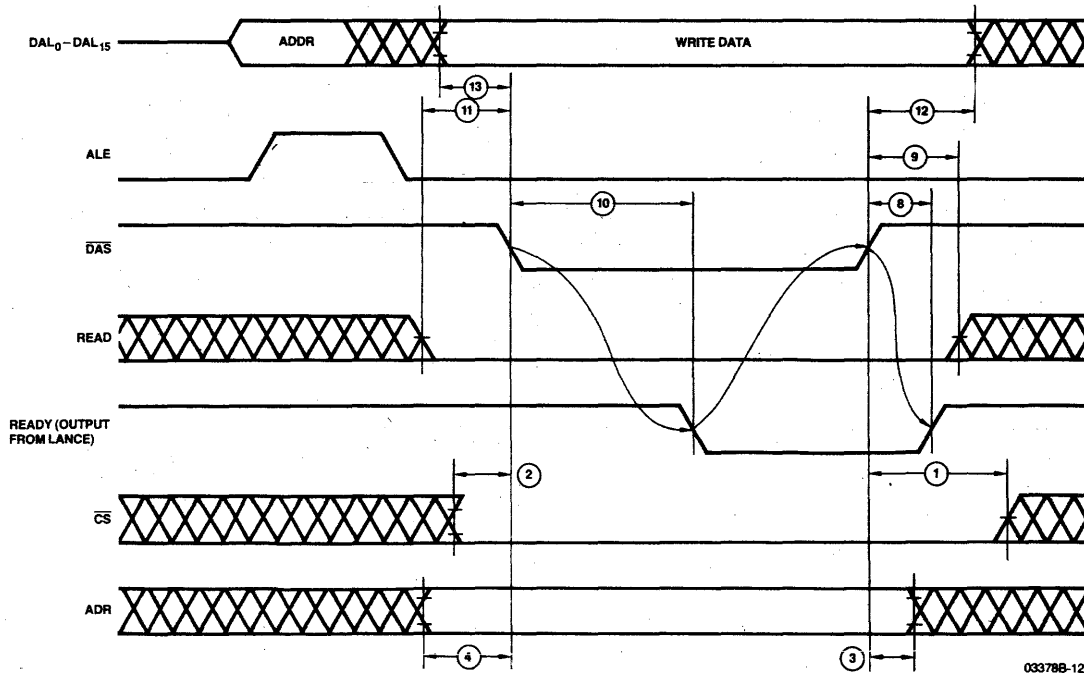
The read cycle is begun by valid addresses being placed on DAL<sub>00</sub>–DAL<sub>15</sub> and A<sub>16</sub>–A<sub>23</sub>. The BYTE MASK signals are placed valid to indicate a word, upper byte or lower byte memory reference and READ indicates the type of cycle. ALE or  $\overline{AS}$  are pulsed and the trailing edge of either can be used to latch addresses. DAL<sub>00</sub>–DAL<sub>15</sub> go into a 3-state mode and  $\overline{DAS}$  falls low to signal the beginning of the memory access. The memory responds by placing  $\overline{READY}$  low to indicate that the DAL lines have valid data. The LANCE then latches memory data on the rising edge of  $\overline{DAS}$  which in turn ends the memory cycle and  $\overline{READY}$  returns high.

The bus transceiver controls,  $\overline{DALI}$  and  $\overline{DALO}$ , are used to control the bus transceivers.  $\overline{DALI}$  signals to strobe data toward the LANCE and  $\overline{DALO}$  signals to strobe data or addresses away from the LANCE. During a read cycle  $\overline{DALO}$  goes inactive before  $\overline{DALI}$  becomes active to avoid "spiking" of the bus transceivers.

### WRITE SEQUENCE

The write cycle is very similar except that the DAL<sub>00</sub>–DAL<sub>15</sub> lines change from containing addresses to data after ALE or  $\overline{AS}$  go inactive.  $\overline{DAS}$  goes active after data is valid on the bus. Data to memory is held valid after  $\overline{DAS}$  goes inactive.

BUS SLAVE READ TIMING



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## COMMUNICATIONS Am7990 Family Ethernet Node

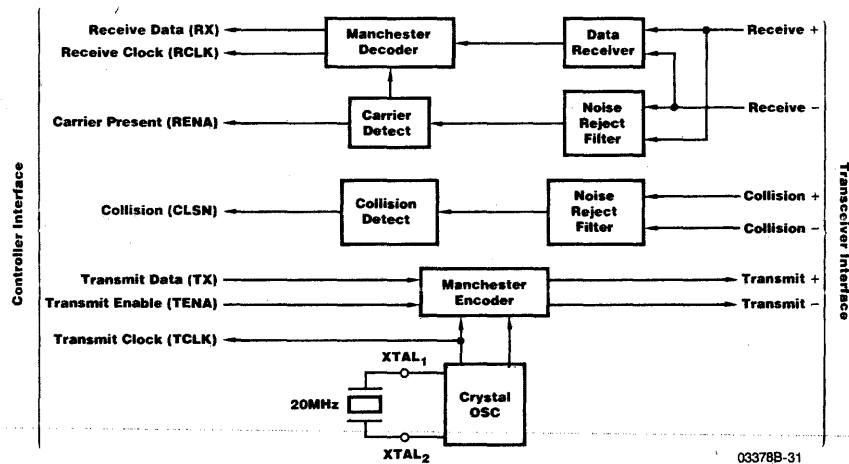
### DISTINCTIVE CHARACTERISTICS

- Compatible with Ethernet/IEEE 802.3 specifications
- Crystal controlled Manchester Encoder
- Manchester Decoder acquires clock and data within six bit times with an accuracy of  $\pm 3\text{ns}$
- Guaranteed carrier detection and collision detection threshold limits
  - Carrier/collision detected for greater than  $-300\text{mV}$
  - No carrier/collision for less than  $-175\text{mV}$
- Input signal conditioning rejects transient noise
  - Transients  $< 10\text{ns}$  for collision detector inputs
  - Transients  $< 16\text{ns}$  for carrier detector inputs
- Receiver decodes Manchester data with up to  $\pm 20\text{ns}$  clock jitter (at 10MHz)
- TTL compatible host interface
- Transmit accuracy  $\pm 0.01\%$  (without adjustments)

### GENERAL DESCRIPTION

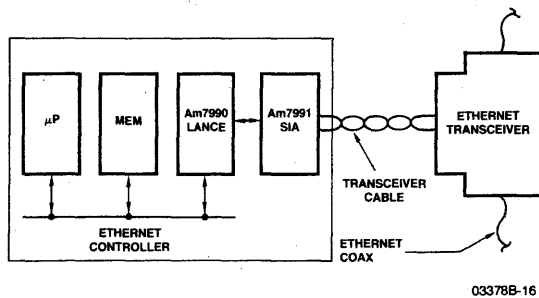
The Am7991 Serial Interface Adapter (SIA) is a Manchester Encoder/Decoder compatible with both Ethernet and IEEE-802.3 specifications. In an Ethernet/IEEE-802.3 application the Am7991 interfaces the Am7990 Local Area Network Controller for Ethernet (LANCE) to the Ethernet transceiver cable, acquires clock and data within 6 bit-times and decodes Manchester data with up to  $\pm 20\text{ns}$  phase jitter at 10MHz. SIA provides both guaranteed signal threshold limits and transient noise suppression circuitry in both data and collision paths to minimize false start conditions.

### BLOCK DIAGRAM Serial Interface Adapter (SIA)



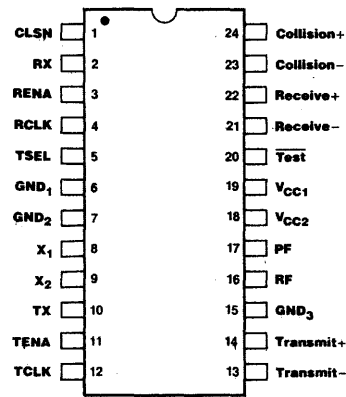
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### TYPICAL ETHERNET NODE



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### CONNECTION DIAGRAM - Top View D-24-1



03378B-32

## DTMF TONE GENERATOR

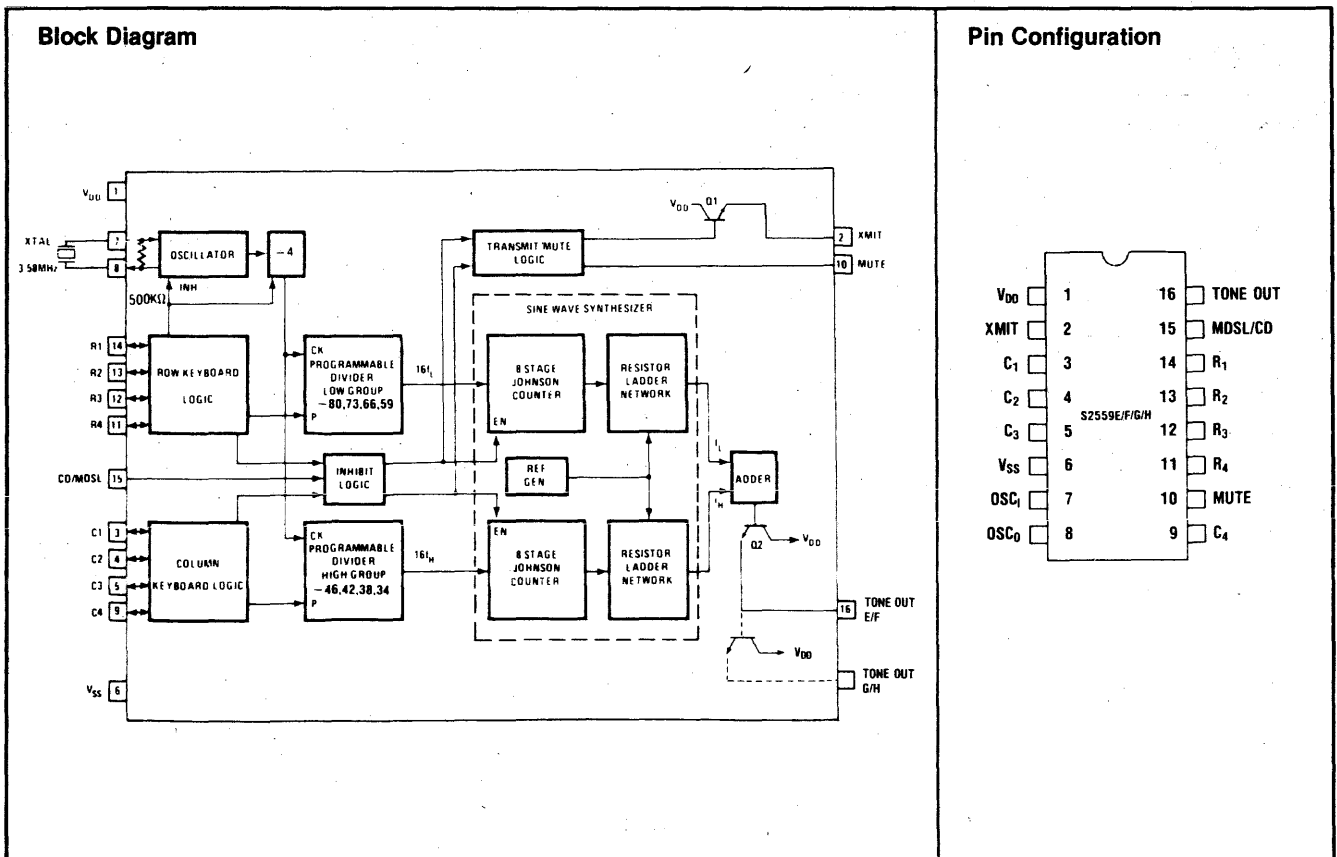
### Features

- Low Output Tone Distortion: 7%
- Wide Operating Supply  
Voltage Range: 2.5 to 10 Volts
- Oscillator Bias Resistor On-Chip
- Can be Powered Directly from Telephone Line or from Small Batteries
- Interfaces Directly to a Standard Telephone Push-Button or Calculator Type X-Y Keyboard
- Four Options Available on Pin 15 and Pin 16:  
Bipolar Output  
E: Mode Select  
F: Chip Disable  
Darlington Output  
G: Mode Select  
H: Chip Disable

### General Description

The S2559E, F, G and H are improved members of the S2559 Tone Generator Family. The new devices feature extended operating voltage range, lower tone distortion, and an on-chip oscillator bias resistor. The S2559E and F are pin and functionally compatible with the S2559C and D, respectively.

The S2559 G and H are identical to the E and F, except that there is a Darlington amplifier configuration on the tone out pin, rather than a single bipolar transistor as shown in the block diagram. In many applications this eliminates the need for a transistor in the telephone circuit. Tone distortion in the telephone is also likely to be lower.





## DTMF TONE GENERATOR

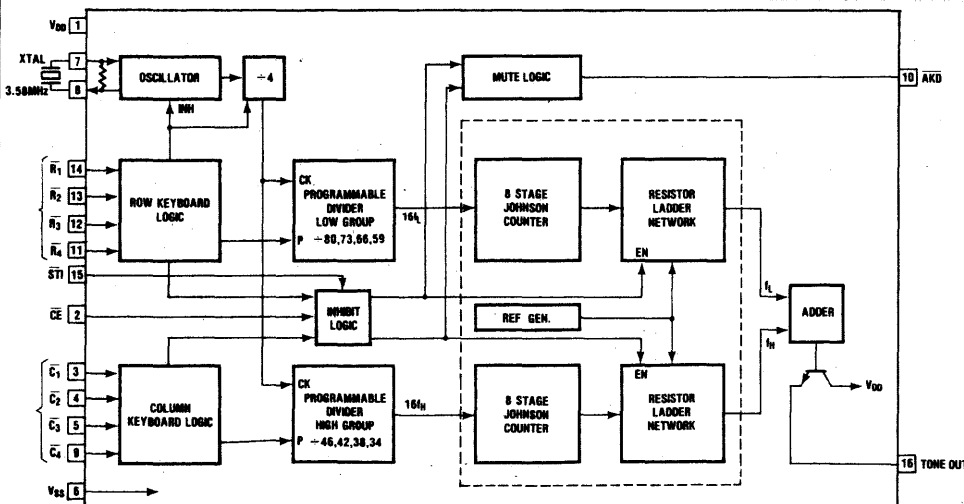
### Features

- Wide Operating Voltage Range: 2.5 to 10 Volts
- Optimized for Constant Operating Supply Voltages, Typically 3.5V
- Tone Amplitude Stability is Within  $\pm 1.5\text{dB}$  of Nominal Over Operating Temperature Range
- Low Power CMOS Circuitry Allows Device Power to be Derived Directly From the Telephone Lines or From Small Batteries
- Uses TV Crystal Standard (3.58MHz) to Derive All Frequencies Thus Providing Very High Accuracy and Stability
- Specifically Designed for Electronic Telephone Applications
- Interfaces Directly to a Standard Telephone Push-Button Keyboard With Common Terminal
- Low Total Harmonic Distortion
- Single Tone as Well as Dual Tone Capability
- Direct Replacement for Mostek MK5089 Tone Generator

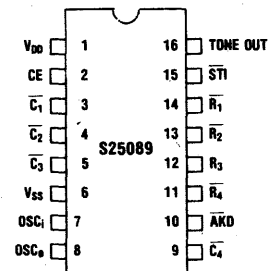
### General Description

The S25089 DTMF Generator is specifically designed to implement a dual tone telephone dialing system in applications requiring fixed supply operation and high stability tone output level, making it well suited for electronic telephone applications. The device can interface directly to a standard pushbutton telephone keyboard with common terminal connected to  $V_{SS}$  and operates directly from the telephone lines. All necessary dual-tone frequencies are derived from the widely used TV crystal standard providing very high accuracy and stability. The required sinusoidal waveform for the individual tones is digitally synthesized on the chip. The waveform so generated has very low total harmonic distortion. A voltage reference is generated on the chip which is very stable over the operating temperature range and regulates the signal levels of the dual tones to meet the recommended telephone industry specifications.

### Block Diagram



### Pin Configuration



## PULSE DIALER

### Features

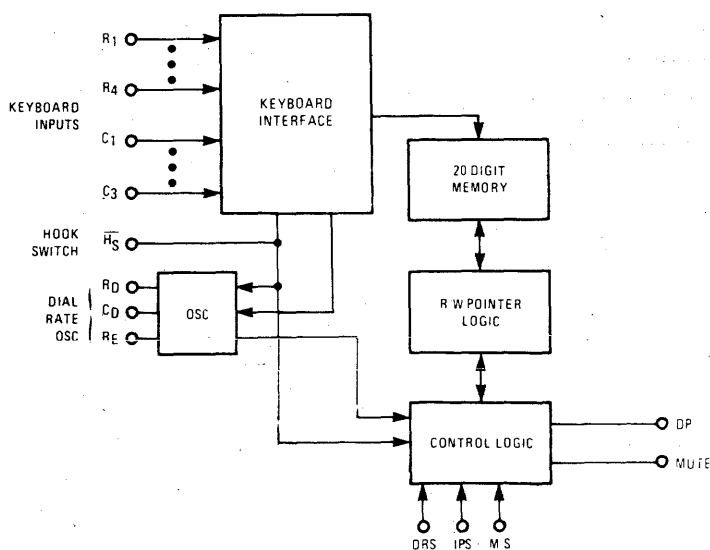
- Low Voltage CMOS Process for Direct Operation from Telephone Lines
- Inexpensive R-C Oscillator Design Provides Better than  $\pm 5\%$  Accuracy Over Temperature and Unit to Unit Variations
- Dialing Rate Can be Varied by Changing the Dial Rate Oscillator Frequency
- Dial Rate Select Input Allows Changing of the Dialing Rate by a 2:1 Factor Without Changing Oscillator Components
- Two Selections of Mark/Space Ratios ( $33\frac{1}{3}/66\frac{2}{3}$  or 40/60)
- Twenty Digit Memory for Input Buffering and for Redial with Access Pause Capability
- Mute and Dial Pulse Drivers on Chip

- Accepts DPCT Keypad with Common Arranged in a 2 of 7 Format; Also Capable of Interface to SPST Switch Matrix

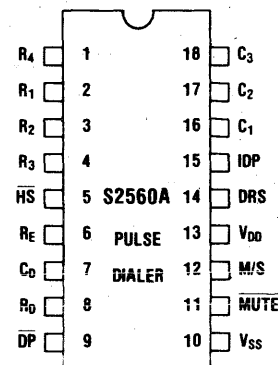
### General Description

The S2560A Pulse Dialer is a CMOS integrated circuit that converts pushbutton inputs to a series of pulses suitable for telephone dialing. It is intended as a replacement for the mechanical telephone dial and can operate directly from the telephone lines with minimum interface. Storage is provided for 20 digits, therefore, the last dialed number is available for redial until a new number is entered. IDP is scaled to the dialing rate such as to produce smaller IDP at higher dialing rates. Additionally, the IDP can be changed by a 2:1 factor at a given dialing rate by means of the IDP select input.

### Block Diagram



### Pin Configuration



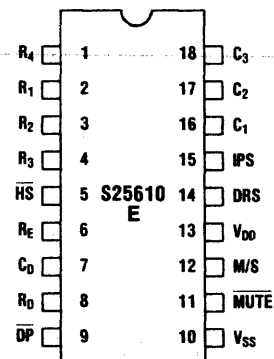
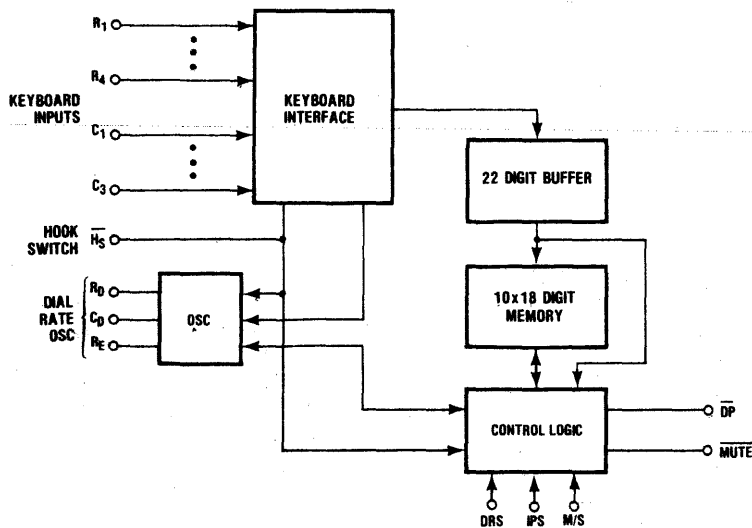


## SINGLE CHIP REPERTORY DIALER

### Features

- Complete Pin Compatibility With S2560A and S2560G Pulse Dialer Allowing Easy Upgrading of Existing Designs.
- Ten 18-Digit Number Memories Plus Last Number Redial (22 Digit) Memory On Chip.
- Low Voltage CMOS Process for Direct Operation From Telephone Lines.
- Inexpensive R-C Oscillator Design With Accuracy Better Than  $\pm 5\%$  Over Temperature and Unit-Unit Variations.
- Independent Select Inputs for Variation of Dialing Rates (10pps/20pps), Mark/Space Ratio ( $33\frac{1}{3}$ - $66\frac{2}{3}$ /40-60), Interdigit Pause (400ms/800ms).
- Can Interface With Inexpensive XY Matrix or Standard 2 of 7 Keyboard With Common. Also Capable of Logic Interface (Active High).
- Mute and Pulse Drivers On Chip.
- Call Disconnect by Pushing \* and # Keys Simultaneously.

### Block Diagram



American Microsystems, Inc. LINEAR



## TONE RINGER

### Features

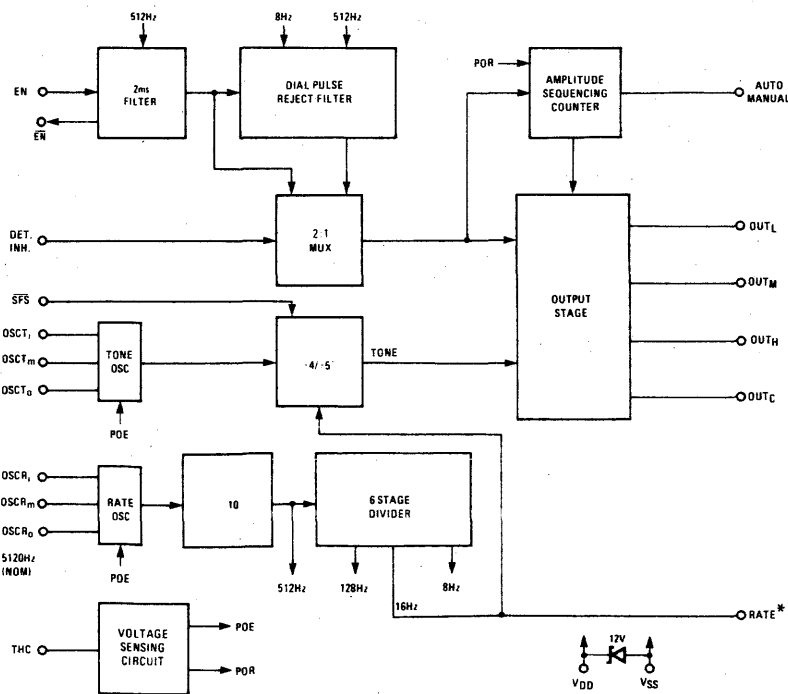
- CMOS Process for Low Power Operation
- Operates Directly from Telephone Lines with Simple Interface
- Also Capable of Logic Interface for Non-Telephone Applications
- Provides a Tone Signal that Shifts Between Two Predetermined Frequencies at Approximately 16Hz to Closely Simulate the Effects of the Telephone Bell
- Push-Pull Output Stage Allows Direct Drive, Eliminating Capacitive Coupling and Provides Increased Power Output
- 50mW Output Drive Capability at 10V Operating Voltage

- Auto Mode Allows Amplitude Sequencing such that the Tone Amplitude Increases in Each of the First Three Rings and Thereafter Continues at the Maximum Level
- Single Frequency Tone Capability

### General Description

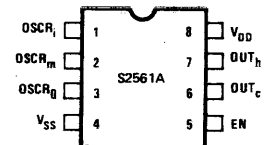
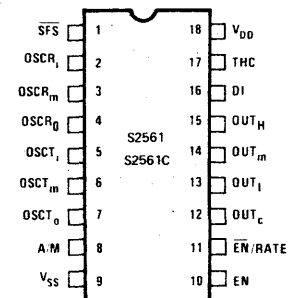
The S2561 Tone Ringer is a CMOS integrated circuit that is intended as a replacement for the mechanical telephone bell. It can be powered directly from the telephone lines with minimum interface and can drive a speaker to produce sound effects closely simulating the telephone bell.

### Block Diagram



\*OPTIONAL OUTPUT FOR S2561C

### Pin Configuration





August 1983

# TWO TO FOUR WIRE TELEPHONE HYBRID WITH TONE RINGER

### Features

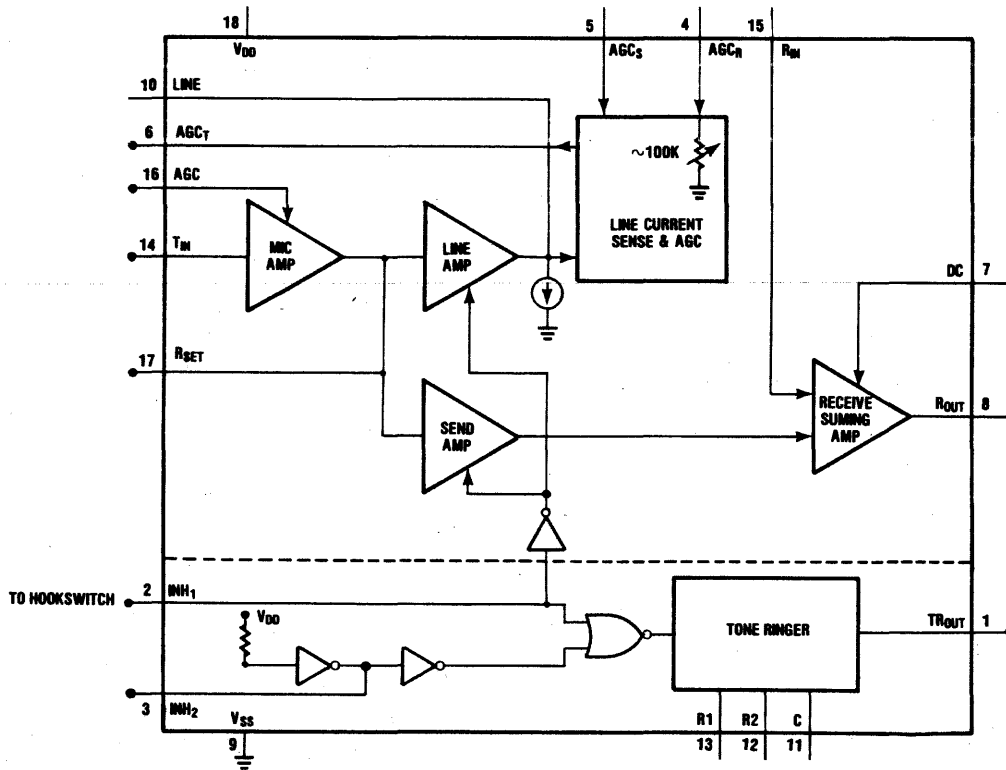
- Monolithic IC Consisting of the Speech Network and Tone Ringer
- Interfaces With Inexpensive Condenser Electret Microphone, Electromagnetic Receiver and a Piezoelectric Ringer Transducer
- Automatic Gain Adjustment for Loop Loss Compensation
- Low Voltage CMOS Process for Operation Over Varying Loop Lengths and Currents

- Uses Inexpensive and Non-Critical External Components

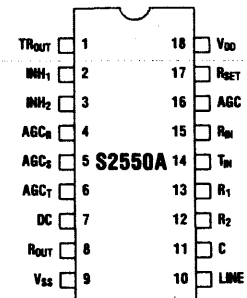
### General Description

The S2550A is a monolithic integrated circuit specifically designed for implementing a low cost telephone set circuit. It consists of the hybrid circuit for speech transmission and reception and a tone ringer circuit that generates an audible tone coincident with the incoming ringing signal through a suitable piezoelectric transducer or high impedance speaker.

### Functional Block Diagram



### Pin Configuration





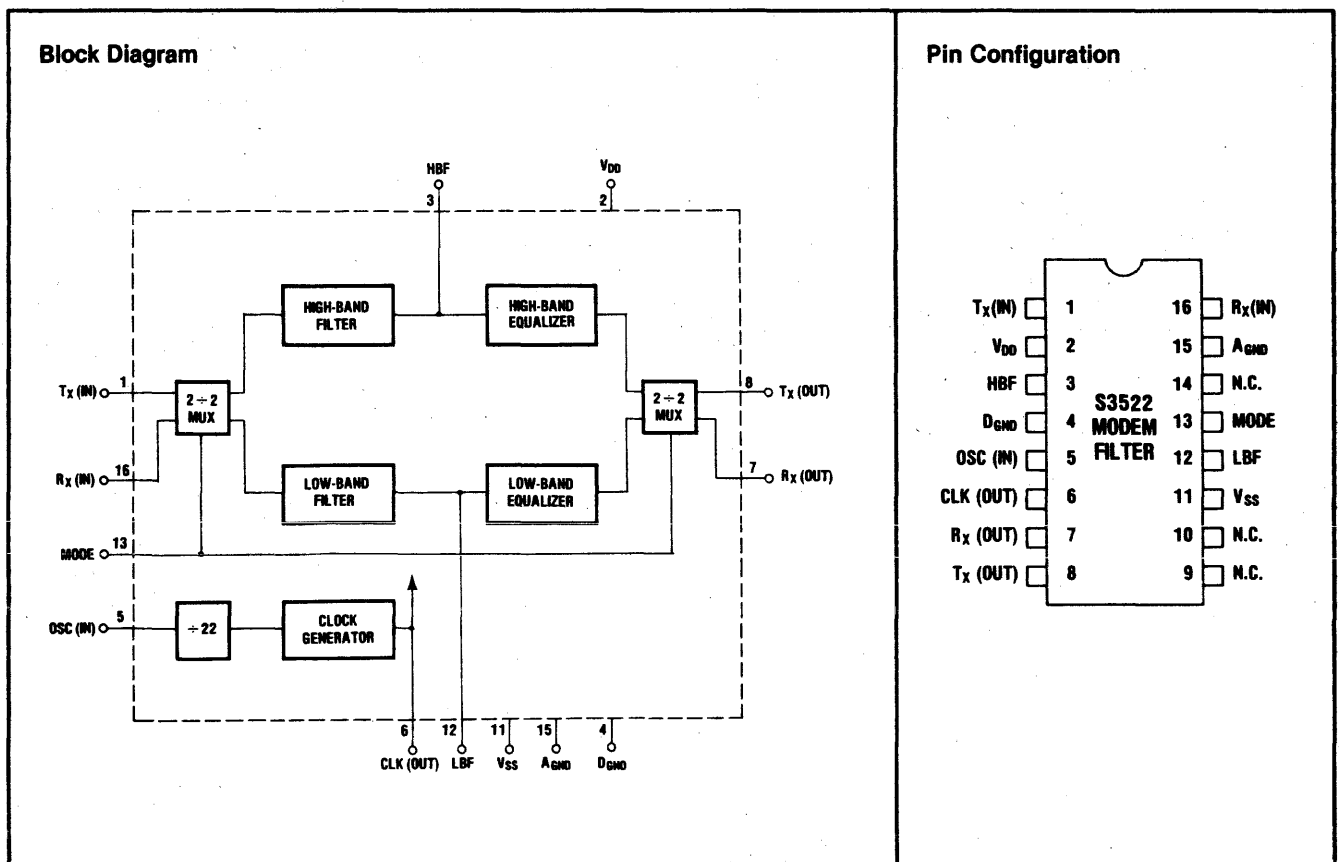
## BELL 212A/CCITT V.22 COMPATIBLE MODEM FILTER WITH EQUALIZER

### Features

- Bell 212A/V.22 Compatible
- Usable for Bell 103/113 Applications
- High and Low Band Filters with Compromise Group Delay Equalizers
- Originate/Answer Operating Modes
- Buffered Clock Output
- Excellent Rejection of CCITT Guard Tones
- Low Power CMOS 50mW Typ.
- ± 5 Volt Operation
- Low Cost 16-Pin Package

### General Description

The AMI S3522 Modem Filter is a 16-pin monolithic CMOS integrated circuit designed to implement both the filtering and equalizing functions required in Bell 212A and CCITT V.22 Modems. The S3522 includes both the transmit signal shaping filter and the receive signal separation filter and features on-chip originate/answer mode selection logic. In addition, half-channel compromise amplitude and group delay equalization is included, giving full compromise equalization through the transmit and receive filter pair. The S3522 features excellent rejection of the CCITT Guard Tones at 550Hz: Low-Band (56dB), High-Band (61dB) and 1800Hz: Low-Band (48dB), High-Band (28dB).







## SINGLE FREQUENCY TUNEABLE BANDPASS/NOTCH FILTER/TONE GENERATOR

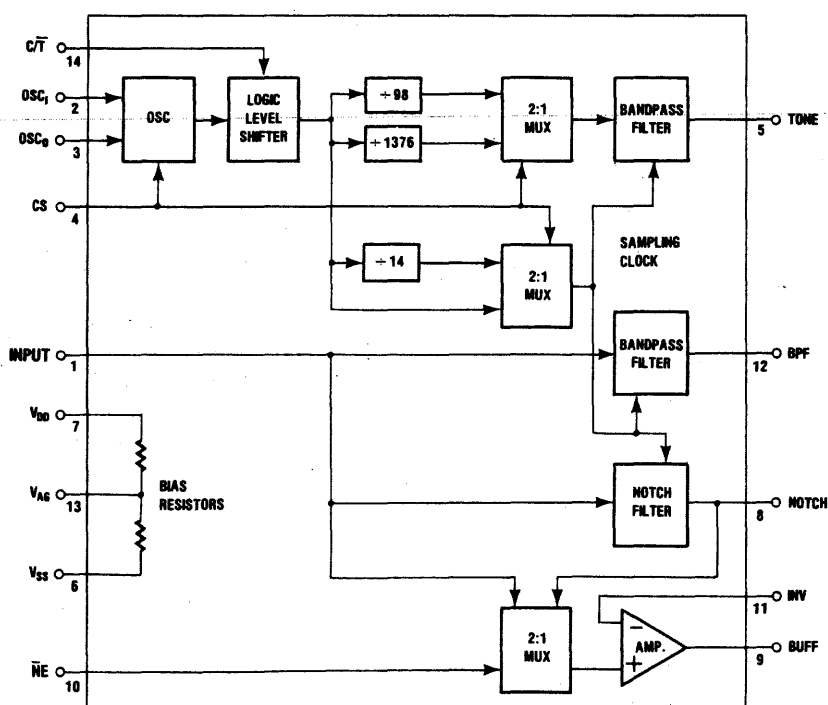
### Features

- Center Frequency of Filters Match and Track Frequency of Generated Tone
- Tone Frequency Adjustable Over a 100Hz to 5kHz Range
- Unfiltered Input, Input with Notched Tone, Input Tone and Tone Generator Outputs
- Operation from a Crystal or External CMOS/TTL Clock
- Operation at 2600Hz from a Low Cost 3.58MHz TV Color Burst Crystal or 256kHz Ext. Clock
- Buffered Output Drives 600Ω Loads
- Single or Split Supply Operation
- Low Power CMOS Technology

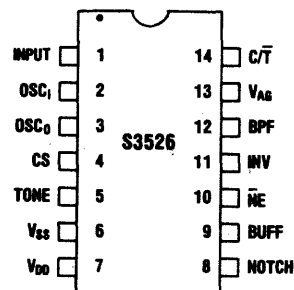
### General Description

The S3526 is a low power CMOS Circuit which may be used in a variety of single frequency (SF) communication applications such as SF-Tone Receivers, Tone Remote Control in Mobile systems, Loopback Diagnostics in Modems, control of Echo Cancellers, dialing and privacy functions in Common Carrier Radio Telephone, etc. The main functional blocks of the S3526 include a low distortion tone (sinewave) generator whose frequency may be programmed using a crystal (i.e. 2600Hz using a low cost TV color burst crystal) or external clock time base; a bandpass filter used to extract tone information from the input signal; a band reject filter which is used to "Notch" out tone information from the input signal; and a buffer amplifier with selectable input (unfiltered input signal, or input signal with tone notched) capable of driving a 600Ω load.

### Block Diagram



### Pin Configuration



## PROGRAMMABLE LOW PASS FILTER

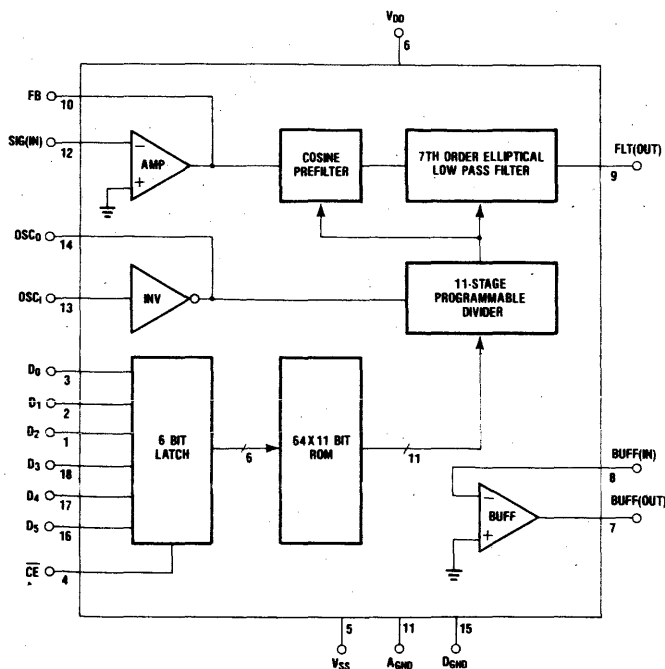
### Features

- Seventh Order Elliptical Ladder Filter with Cosine Prefiltering Stage
- Cutoff Frequency ( $f_c$ ) Range of 10Hz to 20kHz, 40Hz to 20kHz Via 3.58MHz TV Crystal
- Passband Ripple: <0.1dB
- Stopband Attenuation: >51dB for  $f > 1.3f_c$
- Cutoff Frequency Selectable in 64 Steps Via Six Bit Control Word
- Steps May Be Custom Programmed from a Set of 2,048 Discrete Points Via Internal ROM
- Cutoff Frequency Continuously Variable Via External Clock (Crystal, Resonator, or TTL/CMOS Clock)
- Uncommitted Input and Output Op Amps for Anti-Aliasing and Smoothing Functions
- Low Power CMOS Technology

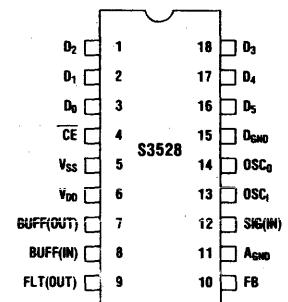
### General Description

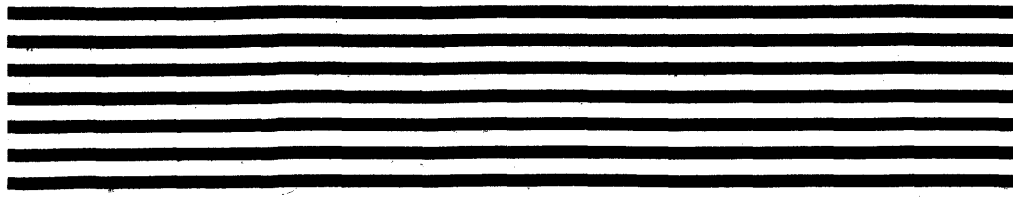
The S3528 is a programmable low pass filter which may be used in a wide variety of filtering applications commonly found in speech analysis, telecommunications, test equipment and instrumentation, etc. The 3528's CMOS design using switched capacitor design techniques allows easy programming of the filter's cutoff frequency ( $f_c$ ), discretely, in 64 steps via a six bit control word or continuously by varying the external time base. The useful range of operation of the filter passband extends from 10Hz to 20kHz. When operating from a low cost TV crystal (3.58MHz) a range of 40Hz to 20kHz may be realized. For special applications the S3528 may be customized via the internal ROM to accommodate a specific set of cutoff frequencies from a choice of 2,048 possibilities.

### Block Diagram



### Pin Configuration





# CMOS SINGLE CHIP $\mu$ -LAW/A-LAW SYNCHRONOUS COMBO CODECS WITH FILTERS

### Features

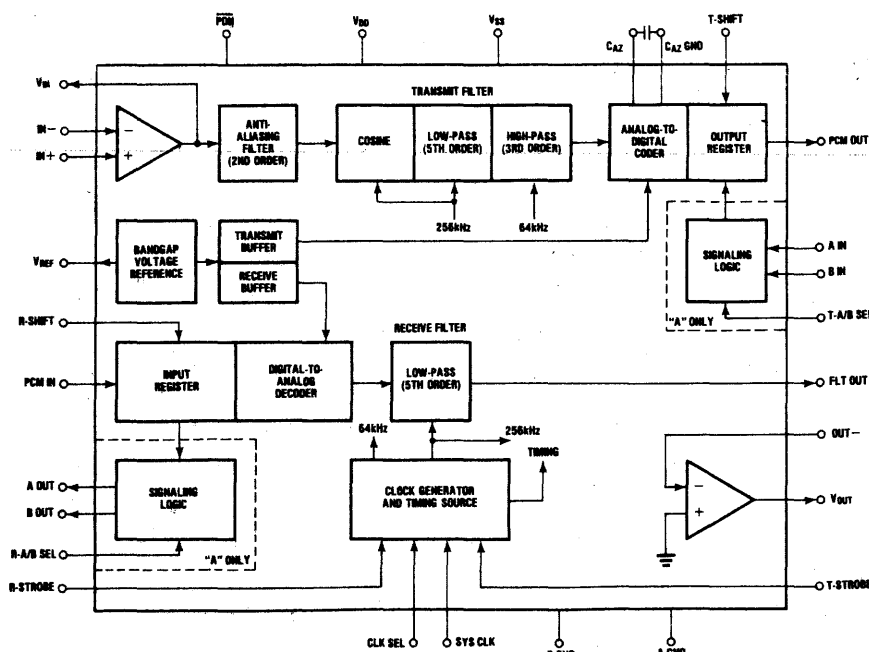
- Independent Transmit and Receive Sections With 75dB Isolation
- Low Power CMOS 80mW (Operating) 8mW (Standby)
- Stable Voltage Reference On-Chip
- Meets or Exceeds AT&T D3, and CCITT G.711, G.712 and G.733 Specifications
- Input Analog Filter Eliminates Need for External Anti-Aliasing Pre-filter
- Input/Output Op Amps for Programming Gain
- Output Op Amp Provides  $\pm 3.1V$  into a 600 $\Omega$  Load or Can Be Switched Off for Reduced Power (70mW)
- Special Idle Channel Noise Reduction Circuitry for Crosstalk Suppression

- Encoder has Dual-Speed Auto-Zero Loop for Fast Acquisition on Power-Up
- Low Absolute Group Delay = 450 $\mu$ sec. @ 1kHz

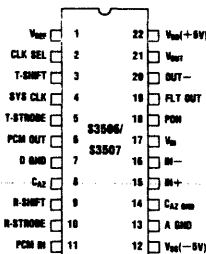
### General Description

The S3506 and S3507 are monolithic silicon gate CMOS Companding Encoder/Decoder chips designed to implement the per channel voice frequency Codecs used in PCM systems. The chips contain the band-limiting filters and the analog  $\leftrightarrow$  digital conversion circuits that conform to the desired transfer characteristic. The S3506 provides the European A-Law companding and the S3507 provides the North American  $\mu$ -Law companding characteristic.

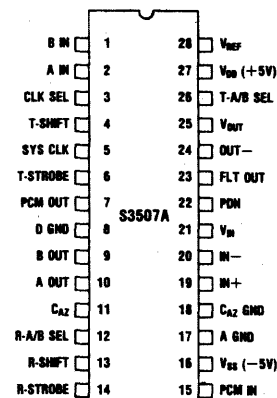
### Block Diagram



### Pin Configuration (22 Pin)



### Pin Configuration (28 Pin)





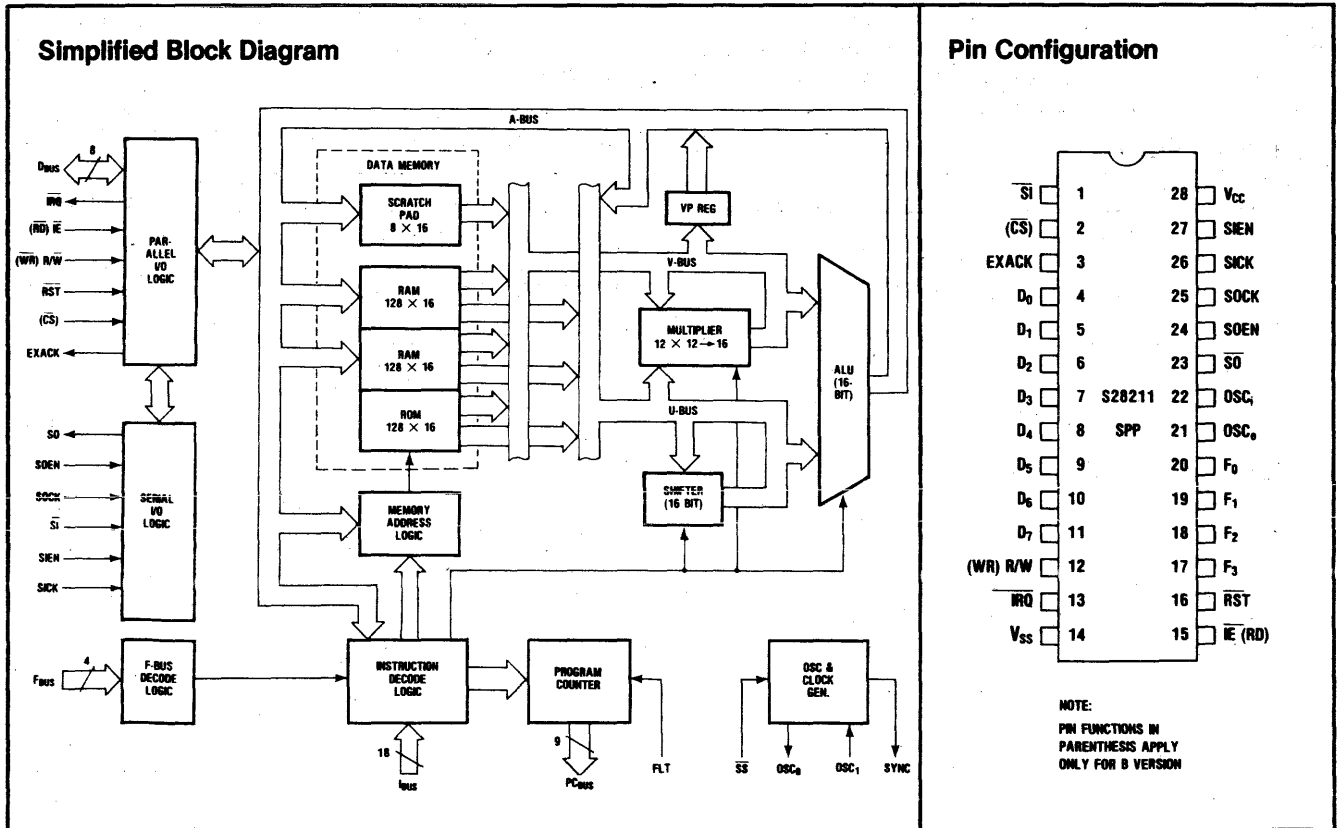
## SIGNAL PROCESSING PERIPHERAL

### Features

- Single-Chip Programmable Digital Signal Processor
- May Be Customized (ROM Programmed) With Customer Generated Routines
- Self-Emulation Capability
- Standard Preprogrammed Processors Available
- Fetch/Multiply/Add/Store Cycle
- 512 Word  $\times$  18 Bit Instruction Memory
- Unique Three Port Data Memory  
256  $\times$  16 RAM/128  $\times$  16 ROM
- 12  $\times$  12 Pipelined Multiplier With 16 Bit Product
- 16 Bit Accumulator With Overflow Detect/Protect
- Double Buffered Asynchronous Serial I/O Port
- $\mu$ P-Compatible I/O Port i.e. 6800 (A Version), 8080 (B Version), etc.

### General Description

The S28211 is a single-chip microcomputer which has been optimized to execute digital signal processing algorithms commonly used in applications such as telecommunications speech processing, industrial process control instrumentation, etc. It may be used as a stand alone unit, or may be operated as a peripheral in a microprocessor based system. The latter configuration allows arrays of S28211s to be used together for increased processing throughput. The S28211's multi-bus, pipelined architecture and powerful multi-operation instructions make it possible to write very compact algorithms. This allows the available memory to be used efficiently and increases the execution speed of a given algorithm. The S28211 may be



## AD547/AD647

### FEATURES

- Ultra Low Drift ( $1\mu\text{V}/^\circ\text{C}$ )
- Low Offset Voltage (0.25mV)
- Low Input Bias Currents (25pA)
- Low Quiescent Current
- Low Noise ( $2\mu\text{V p-p}$ )
- High Open Loop Gain (110dB)
- Matched Offset Voltage – AD647
- Matched Offset Voltage Over Temperature – AD647
- Matched Bias Current – AD647
- Crosstalk 124dB at 1kHz
- Low Total Harmonic Distortion

### PRODUCT DESCRIPTION

The AD547 and AD647 are monolithic, FET input operational amplifiers combining the very low input bias current advantages of a BIFET op amp with offset and drift performance previously available only in high quality bipolar amplifiers.

The exclusive Analog Devices laser wafer trim process trims both the input offset voltage and offset voltage drift to levels far lower than any competing BIFET amplifier ( $1\text{mV}$ ,  $5\mu\text{V}/^\circ\text{C}$ —J versions,  $0.25\text{mV}$ ,  $1\mu\text{V}/^\circ\text{C}$ —L versions).

In addition, the offset voltage is laser trimmed to less than  $0.25\text{mV}$  and matched to  $0.25\text{mV}$  for the AD647L,  $0.5\text{mV}$  and matched to  $0.25\text{mV}$  for the AD647K.

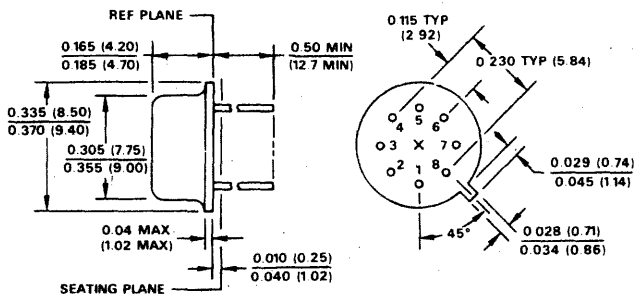
The AD547 offers the lowest guaranteed input bias currents of any BIFET amplifier with  $50\text{pA}$  max for the J grade and  $25\text{pA}$  max for the L grade. The AD647 offers matched bias currents that are significantly lower than currently available monolithic dual FET input operational amplifiers:  $35\text{pA}$  max, matched to  $15\text{pA}$  for the AD647K and L;  $75\text{pA}$  max, matched to  $25\text{pA}$  for the AD647J and S.

### PRODUCT HIGHLIGHTS

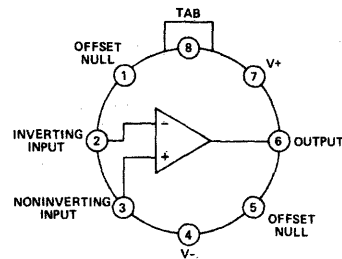
1. Advanced laser wafer trimming techniques reduce offset voltage drift to  $1\mu\text{V}/^\circ\text{C}$  max and reduce offset voltage to only  $0.25\text{mV}$  max on the AD547L.

### OUTLINE DIMENSIONS

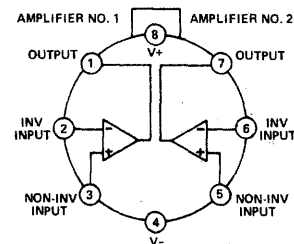
Dimensions shown in inches and (mm).



TO-99



AD547 SINGLE VERSION



AD647 DUAL VERSION

2. Analog Devices BIFET processing provides  $25\text{pA}$  max ( $10\text{pA}$  typical) bias currents specified after 5 minutes of warm-up.
3. Low voltage noise, high open loop gain and outstanding offset performance make the series true precision BIFET amplifiers.
4. The AD647 dual has tight matching specifications to ensure high performance, eliminating the need to match individual devices.
5. Laser-wafer-trimming reduces offset voltage to as low as  $0.25\text{mV}$  max and matched side to side to  $0.25\text{mV}$  (AD647L), thus eliminating the need for external nulling.
6. The standard dual amplifier pin out allows the AD647 to replace lower performance duals without redesign.

### ORDERING GUIDE

Model	Initial Offset Voltage	Offset Voltage Drift	Temperature Range
AD547JH	1.0mV	$5\mu\text{V}/^\circ\text{C}$	0 to $+70^\circ\text{C}$
AD547KH	0.5mV	$2\mu\text{V}/^\circ\text{C}$	0 to $+70^\circ\text{C}$
AD547LH	0.25mV	$1\mu\text{V}/^\circ\text{C}$	0 to $+70^\circ\text{C}$
AD547SH	0.5mV	$5\mu\text{V}/^\circ\text{C}$	$-55^\circ\text{C}$ to $+125^\circ\text{C}$
AD647JH	1.0mV	$10\mu\text{V}/^\circ\text{C}$	0 to $+70^\circ\text{C}$
AD647KH	0.5mV	$5\mu\text{V}/^\circ\text{C}$	0 to $+70^\circ\text{C}$
AD647LH	0.25mV	$2\mu\text{V}/^\circ\text{C}$	0 to $+70^\circ\text{C}$
AD647SH	0.5mV	$10\mu\text{V}/^\circ\text{C}$	$-55^\circ\text{C}$ to $+125^\circ\text{C}$

**FEATURE SELECTION CHART**

		GENERAL PURPOSE										HIGH ACCURACY						
		FET INPUT					WIDEBAND					LOW V <sub>OS</sub> DRIFT						
		AD103	AD206	AD310	AD542	AD642	AD644	AD744	AD101 Series	AD741	AD104	AD310	AD OP-07	AD517	AD545	AD547	AD647	AD715
Monolithic Technology	Bipolar Input J-FET Dual J-FET			•	•	•	•	•	•	•	•	•	•		•	•		
Multi-Device Technology	Hybrid Module	•	•											•			•	
High Open Loop Gain	>100dB >140dB				•	•					•	•	•		•	•		
High CMR	>100dB										•	•	•					
Low Offset Voltage	≤5mV ≤1mV ≤50μV		•		•	•	•	•	•	•	•	•	•	•	•	•	•	
Low Offset V. vs. Temp	≤5μV/°C ≤1μV/°C ≤0.6μV/°C				•	•	•	•	•	•	•	•	•	•	•	•	•	
Low Bias Current	≤50pA ≤5pA ≤0.5pA	•	•	•	•	•	•	•	•	•				•	•	•	•	
Fast Settling	≤1μs to 0.1% ≤5μs to 0.01%						•	•										
Wideband (Unity Gain)	≥2MHz ≥10MHz						•	•	•									
High Slew Rate	>10V/μs >30V/μs >100V/μs >1000V/μs						•	•										
Low Noise (0.1 to 10Hz)	2μV p-p				•	•	•	•	•	•	•	•	•					
High Voltage Out High Current Out Low Power	>100V >20mA ≤75mW				•	•	•	•	•	•	•	•	•		•	•	•	
Second Source								•	•			•						
Temperature Range	0 to +70°C -25°C to +85°C -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	
Dice Availability				•	•	•	•	•	•	•	•	•	•					

**GENERAL PURPOSE BIPOLAR INPUT SPECIFICATIONS (typical @ V<sub>S</sub> = ±15V, T<sub>A</sub> = +25°C unless otherwise noted)**

Model	Gain min	Output V/mA min	Unity Gain MHz	Slew Rate V/μs	V <sub>OS</sub> mV max	ΔV <sub>OS</sub> /ΔT μV/°C max	I <sub>b</sub> nA max	CMR dB min	Temp Range <sup>1</sup>	Package <sup>2</sup>
AD101A	50k	10/5	1.0 to 8.0	0.5 to 10	2.0	15	75	80	M	H
AD201A	50k	10/5	1.0 to 8.0	0.5 to 10	2.0	15	75	80	I	H,N
AD301A	25k	10/5	1.0 to 8.0	0.5 to 10	7.5	30	250	70	C	H,N
AD301AL	80k	10/5	1.0 to 10.0	0.25 to 9.0	1.0	5.0	30	90	C	H,N
AD108	50k	13/1.3	0.3 to 3.0	0.3 to 1.3	2.0	15	2.0	85	M	H
AD208	50k	13/1.3	0.3 to 3.0	0.3 to 1.3	2.0	15	2.0	85	I	H
AD308	25k	13/1.3	0.3 to 3.0	0.3 to 1.3	7.5	30	7.0	80	C	H
AD108A	80k	13/1.3	0.3 to 3.0	0.3 to 1.3	0.5	5.0	2.0	96	M	H
AD208A	80k	13/1.3	0.3 to 3.0	0.3 to 1.3	0.5	5.0	2.0	96	I	H
AD308A	80k	13/1.3	0.3 to 3.0	0.3 to 1.3	0.5	5.0	7.0	96	C	H
AD741	50k	10/5	1.0	0.5	5.0	—	500	70	M	H,N
AD741C	20k	10/5	1.0	0.5	6.0	—	500	70	C	H,N
AD741J	50k	10/10	1.0	0.5	3.0	20	200	80	C	H,N
AD741K	50k	10/5	1.0	0.5	2.0	15	75	90	C	H,N
AD741L	50k	10/5	1.0	0.5	0.5	5.0	50	90	C	H,N
AD741S	50k	10/10	1.0	0.5	2.0	15	75	80	M	H

**NOTES**
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -55°C to +125°C

<sup>2</sup>H = TO-99 can, N = 8-pin mini DIP

## GENERAL PURPOSE FET INPUT

SPECIFICATIONS (typical @  $V_S = \pm 15V$ ,  $T_A = +25^\circ C$  unless otherwise noted)

Model	Gain min	Output V/mA min	Unity Gain MHz	Slew Rate V/ $\mu$ s	$V_{OS}$ mV max	$\Delta V_{OS}/\Delta T$ $\mu V/^\circ C$ max	$I_b$ pA max	CMR dB min	Temp Range <sup>1</sup>	Package <sup>2</sup>
AD503J	20k	10/5	1.0	3.0	50	75	15	70	C	II
AD503K	50k	10/5	1.0	3.0	20	25	10	80	C	II
AD503S	50k	10/5	1.0	3.0	20	50	10	80	M	II
AD506J	20k	10/5	1.0	3.0	3.5	75	15	70	C	II
AD506K	50k	10/5	1.0	3.0	1.5	25	10	80	C	II
AD506L	75k	10/5	1.0	3.0	1.0	10	5	80	C	II
AD506S	50k	10/5	1.0	3.0	1.5	50	10	80	M	II
AD540J	20k	10/5	1.0	6.0	50	75	50	70	C	II
AD540K	50k	10/5	1.0	6.0	20	25	25	70	C	II
AD540S	50k	10/5	1.0	6.0	20	50	25	70	M	II
AD542J	100k	10/5	1.0	3.0	2.0	20	50	76	C	II
AD542K	300k	10/5	1.0	3.0	1.0	10	25	80	C	II
AD542L	300k	10/5	1.0	3.0	0.5	5	25	80	C	II
AD542S	300k	10/5	1.0	3.0	1.0	15	25	80	M	II
AD544J	30k	10/5	2.0	13	2.0	20	50	74	C	II
AD544K	50k	10/5	2.0	13	1.0	10	25	80	C	II
AD544L	50k	10/5	2.0	13	0.5	5	25	80	C	II
AD544S	50k	10/5	2.0	13	1.0	15	25	80	M	II
AD547J	100k	10/5	1.0	3.0	1.0	5	50	76	C	II
AD547K	250k	10/5	1.0	3.0	0.5	2	25	80	C	II
AD547L	250k	10/5	1.0	3.0	0.25	1	25	80	C	II
AD547S	250k	10/5	1.0	3.0	0.5	5	25	80	M	II
AD642J	100k	10/5	1.0	3.0	2.0 <sup>3</sup>	3.5mV <sup>4</sup>	75	76	C	II
AD642K	300k	10/5	1.0	3.0	1.0 <sup>3</sup>	2.0mV <sup>4</sup>	35	80	C	II
AD642L	300k	10/5	1.0	3.0	0.5 <sup>3</sup>	1.0mV <sup>4</sup>	35	80	C	II
AD642S	300k	10/5	1.0	3.0	1.0 <sup>3</sup>	3.5mV <sup>4</sup>	75	80	M	II
AD644J	30k	10/5	2.0	13	2.0 <sup>3</sup>	3.5mV <sup>4</sup>	75	76	C	II
AD644K	50k	10/5	2.0	13	1.0 <sup>3</sup>	2.0mV <sup>4</sup>	35	80	C	II
AD644L	50k	10/5	2.0	13	0.5 <sup>3</sup>	1.0mV <sup>4</sup>	35	80	C	II
AD644S	50k	10/5	2.0	13	1.0 <sup>3</sup>	3.5mV <sup>4</sup>	75	80	M	II
AD647JH	100k	10/5	1.0	3.0	1.0 <sup>3</sup>	10	75	76	C	II
AD647KH	250k	10/5	1.0	3.0	0.5 <sup>3</sup>	5	35	80	C	II
AD647LH	250k	10/5	1.0	3.0	0.25 <sup>3</sup>	2.5	35	80	C	II
AD647SH	250k	10/5	1.0	3.0	0.5 <sup>3</sup>	10	35	80	M	II

NOTES

<sup>1</sup>C = 0 to +70°C; M = -55°C to +125°C

<sup>2</sup>H = TO-99 can; Z = Module

<sup>3</sup>Dual amplifiers with matched side to side offset voltage

<sup>4</sup>Guaranteed max offset error from  $T_{min}$  to  $T_{max}$

## HIGH ACCURACY LOW $V_{OS}$ DRIFT

SPECIFICATIONS (typical @  $V_S = \pm 15V$ ,  $T_A = +25^\circ C$  unless otherwise noted)

Model	Gain min	Output V/mA min	Unity Gain MHz	Slew Rate V/ $\mu$ s	$V_{OS}$ mV max	$\Delta V_{OS}/\Delta T$ $\mu V/^\circ C$ max	$I_b$ nA max	CMR dB min	Temp Range <sup>1</sup>	Package <sup>2</sup>
AD504J	250k	10/5	0.3 to 1.0 <sup>3</sup>	0.12 to 2.5 <sup>3</sup>	2.5	5.0	200	94	C	H
AD504K	500k	10/5	0.3 to 1.0 <sup>3</sup>	0.12 to 2.5 <sup>3</sup>	1.5	3.0	100	100	C	H
AD504L	1000k	10/5	0.3 to 1.0 <sup>3</sup>	0.12 to 2.5 <sup>3</sup>	0.5	1.0	80	110	C	H
AD504M	1000k	10/5	0.3 to 1.0 <sup>3</sup>	0.12 to 2.5 <sup>3</sup>	0.5	0.5	80	110	C	H
AD504S	1000k	10/5	0.3 to 1.0 <sup>3</sup>	0.12 to 2.5 <sup>3</sup>	0.5	1.0	80	110	M	H
AD510J	250k	10/5	0.3	0.1	0.1	3.0	25	94	C	H
AD510K	1000k	10/5	0.3	0.1	0.05	1.0	13	110	C	H
AD510L	1000k	10/5	0.3	0.1	0.025	0.5	10	110	C	H
AD510S	1000k	10/5	0.3	0.1	0.05	1.0	13	110	M	H
AD517J	1000k	10/10	0.25	0.1	0.150	3.0	5	94	C	H
AD517K	1000k	10/10	0.25	0.1	0.075	1.8	2	110	C	H
AD517L	1000k	10/10	0.25	0.1	0.05	1.3	1	110	C	H
AD517S	1000k	10/10	0.25	0.1	0.075	1.8	2	110	M	H
AD547J	100k	10/5	1.0	3.0	1.0	5	30	76	C	H
AD547K	250k	10/5	1.0	3.0	0.5	2	25	80	C	H
AD547L	250k	10/5	1.0	3.0	0.25	1	25	80	C	H
AD547S	250k	10/5	1.0	3.0	0.5	5	25	80	M	H
AD647JH	100k	10/5	1.0	3.0	1.0 <sup>3</sup>	10	75	76	C	H
AD647KH	250k	10/5	1.0	3.0	0.5 <sup>3</sup>	5	35	80	C	H
AD647LH	250k	10/5	1.0	3.0	0.25 <sup>3</sup>	2.5	35	80	C	H
AD647SH	250k	10/5	1.0	3.0	0.5 <sup>3</sup>	10	35	80	M	H
AD OP-07EH	2000k	10/10	0.6	0.17	0.075	1.3	34	106	C	H
AD OP-07CH	1200k	10/10	0.6	0.17	0.150	1.6	27	100	C	H
AD OP-07DH	1200k	10/10	0.6	0.17	0.150	2.5	212	94	C	H
AD OP-07AH	3000k	10/10	0.6	0.17	0.025	0.6	42	110	M	H
AD OP-07H	2000k	10/10	0.6	0.17	0.075	1.3	33	110	M	H

NOTES

<sup>1</sup>C = 0 to +70°C; M = -55°C to +125°C

<sup>2</sup>H = TO-99 can; N = 8-pin mini DIP

<sup>3</sup>Dependent on compensation capacitor selection

## HIGH ACCURACY LOW BIAS CURRENT

SPECIFICATIONS (typical @  $V_S = \pm 15V$ ,  $T_A = +25^\circ C$  unless otherwise noted)

Model	Gain min	Output V/mA min	Unity Gain MHz	Slew Rate V/ $\mu$ s	$V_{OS}$ mV max	$\Delta V_{OS}/\Delta T$ $\mu V/^\circ C$ max	$I_b$ pA max	CMR dB min	Temp Range <sup>1</sup>	Package <sup>2</sup>
AD515J	20k	10/5	0.35	0.3	3.0	50	0.300	66	C	H
AD515K	40k	10/5	0.35	0.3	1.0	15	0.150	80	C	H
AD515L	25k	10/5	0.35	0.3	1.0	25	0.075	70	C	H
AD545J	20k	10/5	0.7	1.0	1.0	25	2	66	C	H
AD545K	40k	10/5	0.7	1.0	1.0	15	1	70	C	H
AD545L	40k	10/5	0.7	1.0	0.5	5	1	76	C	H
AD545M	40k	10/5	0.7	1.0	0.25	3	1	76	C	H

NOTES

<sup>1</sup>C = 0 to +70°C

<sup>2</sup>H = TO-99 can; Z = Module

<sup>3</sup>Adjustable to zero

## FEATURE SELECTION CHART

		FAST/WIDEBAND											
		FET INPUT							UNITY GAIN BUFFER				
		AD507	AD509	AD518	AD530	AD581	AD582	AD554	HOS-050	HOS-060	ADLH0032	ADLH0033	HOS-100
Monolithic Technology	Bipolar Input J-FET Dual J-FET	•	•	•									
Multi-Device Technology	Hybrid Module				•	•	•	•	•	•	•	•	•
High Open Loop Gain	≥100dB ≥140dB	•											
High CMR	>100dB												
Low Offset Voltage	≤5mV ≤1mV ≤50μV	•	•		•	•	•	•		•	•	•	•
Low Offset V. vs. Temp	≤5μV/°C ≤1μV/°C ≤0.6μV/°C					•	•	•			•	•	•
Low Bias Current	≤50pA ≤5pA ≤0.5pA					•	•	•			•	•	•
Fast Settling	≤1μs to 0.1% ≤5μs to 0.01%	•	•	•		•	•	•	•	•	•	•	•
Wideband (Unity Gain)	≥2MHz ≥10MHz ≥50MHz	•	•	•		•	•	•	•	•	•	•	•
High Slew Rate	≥10V/μs ≥30V/μs ≥100V/μs ≥1000V/μs	•	•	•		•	•	•	•	•	•	•	•
Low Noise (0.1 to 10Hz)	≥2μV p-p					•	•						
High Voltage Out	≥100V												
High Current Out	≥20mA												
Low Power	≤75mW												
Second Source		•	•					•			•	•	
Temperature Range	0 to +70°C -25°C to +85°C -55°C to +125°C	•	•	•	•	•	•	•	•	•	•	•	•
Dice Availability				•									

### FAST/WIDEBAND

SPECIFICATIONS (typical @  $V_S = \pm 15V$ ,  $T_A = +25^\circ C$  unless otherwise noted)

Model	Gain min	Output V/mA min	Unity Gain MHz	Slew Rate V/μs min	Settling Time μs to 0.1%	$V_{OS}$ mV max	$\Delta V_{OS}/\Delta T$ μV/°C max	$I_B$ nA max	CMR dB min	Temp Range <sup>1</sup>	Package <sup>2</sup>
AD380JH	25k	10/50	40	200	0.13	2.0	50	100	60	C	X
AD380KH	25k	10/50	40	200	0.13	1.0	20	100	60	C	X
AD380LH	25k	10/50	40	200	0.13	1.0	10	100	60	C	X
AD380SH	25k	10/50	40	200	0.13	1.0	50	100	60	M	X
AD381JH	60k	10/5	5	20	0.7	1.0	15	100	76	C	H
AD381KH	100k	10/5	5	20	0.7	0.5	10	100	80	C	H
AD381LH	100k	10/5	5	20	0.7	0.25	5	100	80	C	H
AD381SH	100k	10/5	5	20	0.7	1.0	10	100	80	M	H
AD382JH	25k	10/50	5	20	1.3 <sup>3</sup>	1.0	15	100	76	C	X
AD382KH	35k	10/50	5	20	1.3 <sup>4</sup>	0.5	10	50	80	C	X
AD382LH	35k	10/50	5	20	1.3 <sup>4</sup>	0.25	5	50	80	C	X
AD382SH	35k	10/50	5	20	1.3 <sup>4</sup>	1.0	10	50	80	M	X
AD507J	80k	10/10	35	20	0.9	5	15	25	74	C	H
AD507K	100k	10/10	35	25	0.9	3	15	15	80	C	H
AD507S	100k	10/10	35	20	0.9	4	20	15	80	M	H
AD509J	7.5k	10/5	20	80	0.2	10	20	250	74	C	H
AD509K	10k	10/5	20	80	0.2	8	30	200	80	C	H
AD509S	10k	10/5	20	100	0.5	8	30	200	80	M	H
AD518J	25k	10/10	12	50	0.8	10	10	500	70	C	H
AD518K	50k	10/10	12	50	0.8	4	15	250	80	C	H
AD518S	50k	10/10	12	50	0.8	4	20	250	80	M	H
AD554AM	31.6k	10/100	90	1000	0.12	2.0	50	50	60	I	Y
AD554BM	31.6k	10/100	90	1000	0.12	1.0	15	50	60	I	Y
AD554SM	31.6k	10/100	90	1000	0.12	1.0	25	50	60	M	Y
ADLH0032CG	1k	12/10	70	500	0.3	5	25	0.025	60	I	X
ADLH0032G	1k	12/10	70	500	0.3	2	25	0.01	60	M	X
ADLH0033CG	0.98 <sup>5</sup>	12/100	100 <sup>6</sup>	1400	3.2ms <sup>7</sup>	12	50	0.05	-	I	X
ADLH0033G	0.98 <sup>5</sup>	12/100	100 <sup>6</sup>	1500	2.9ms <sup>7</sup>	5	50	0.05	-	M	X
HOS-050	100k	10/100	100	300	0.1	35	150	2	70	M	X
HOS-050A	100k	10/100	100	300	0.1	15	35	2	70	M	X
HOS-050C	100k	10/100	100	300	0.1	65	200	2*	70	I	X
HOS-060SH	100k	10/100	100	300	0.1	1.0	10	2*	70	M	X
HOS-100AH	0.96 <sup>5</sup>	12/100	125 <sup>8</sup>	1400	2.0ms <sup>7</sup>	10	25	5μA	-	I	X
HOS-100SH	0.97 <sup>5</sup>	12/100	125 <sup>8</sup>	1500	2.0ms <sup>7</sup>	5	25	5μA	-	M	X

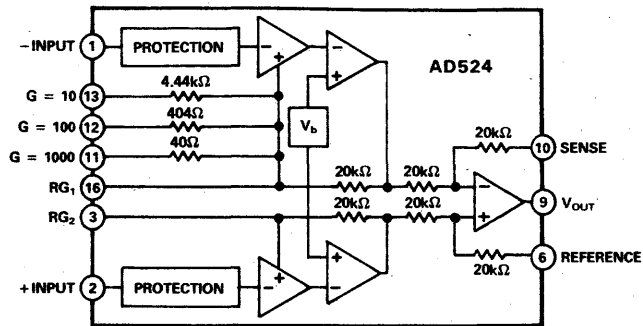
NOTES  
<sup>1</sup>C = 0 to +70°C, I = -25°C to +85°C, M = -25°C to +125°C  
<sup>2</sup>H = TO-99 can, X = 12-lead TO-8 can, Y = TO-3 can, Z = Module  
<sup>3</sup>Settling time to 0.01%  
<sup>4</sup>Settling time to 0.01%, max  
<sup>5</sup>Unity Gain Buffer  
<sup>6</sup>Large Signal Slew Rate  
<sup>7</sup>Rise Time



### FEATURES

- Low Noise:  $0.3\mu\text{V}$  p-p 0.1Hz to 10Hz
- Low Nonlinearity: 0.003% ( $G = 1$ )
- High CMRR: 120dB ( $G = 1000$ )
- Low Offset Voltage:  $50\mu\text{V}$
- Low Offset Voltage Drift:  $0.5\mu\text{V}/^\circ\text{C}$
- Gain Bandwidth Product: 25MHz
- Pin Programmable Gains of 1, 10, 100, 1000
- Input Protection, Power On - Power Off
- No External Components Required
- Internally Compensated

### AD524 FUNCTIONAL BLOCK DIAGRAM



### PRODUCT HIGHLIGHTS

1. The AD524 has guaranteed low offset voltage, offset voltage drift and low noise for precision high gain applications.
2. The AD524 is functionally complete with pin programmable gains of 1, 10, 100 and 1000, and single resistor programmable for any gain.
3. Input and output offset nulling terminals are provided for very high precision applications and to minimize offset voltage changes in gain ranging applications.
4. The AD524 is input protected for both power on and power off fault conditions.
5. The AD524 offers superior dynamic performance with a gain bandwidth product of 25MHz, full power response of 75kHz and a settling time of  $15\mu\text{s}$  to 0.01% of a 10V step ( $G = 100$ ).

### PRODUCT DESCRIPTION

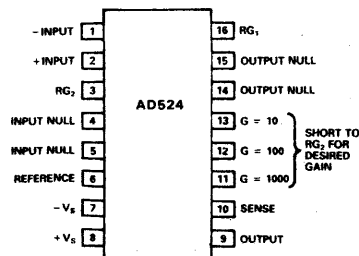
The AD524 is a precision monolithic instrumentation amplifier designed for data acquisition applications requiring high accuracy under worst-case operating conditions. An outstanding combination of high linearity, high common mode rejection, low offset voltage drift, and low noise makes the AD524 suitable for use in many data acquisition systems.

The AD524 has an output offset voltage drift of less than  $25\mu\text{V}/^\circ\text{C}$ , input offset voltage drift of less than  $0.5\mu\text{V}/^\circ\text{C}$ , CMR above 90dB at unity gain (120dB at  $G = 1000$ ) and maximum nonlinearity of 0.003% at  $G = 1$ . In addition to the outstanding dc specifications the AD524 also has a 25MHz gain bandwidth product ( $G = 100$ ). To make it suitable for high speed data acquisition systems the AD524 has an output slew rate of  $5\text{V}/\mu\text{s}$  and settles in  $15\mu\text{s}$  to 0.01% for gains of 1 to 100.

As a complete amplifier the AD524 does not require any external components for fixed gains of 1, 10, 100 and 1,000. For other gain settings between 1 and 1000 only a single resistor is required. The AD524 input is fully protected for both power on and power off fault conditions.

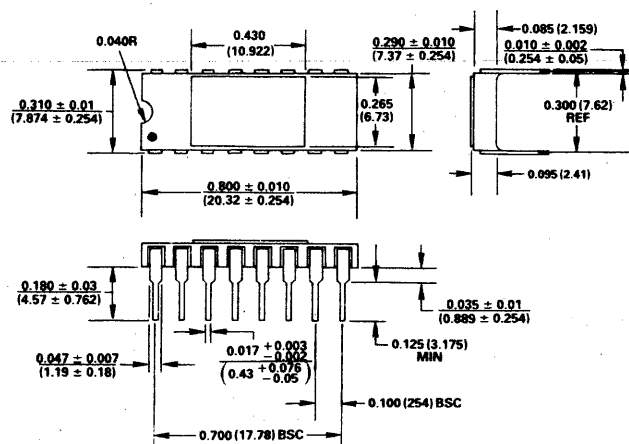
The AD524 IC instrumentation amplifier is available in four different versions of accuracy and operating temperature range. The economical "A" grade, the low drift "B" grade and lower drift, higher-linearity "C" grade are specified from  $-25^\circ\text{C}$  to  $+85^\circ\text{C}$ . The "S" grade guarantees performance to specification over the extended temperature range  $-55^\circ\text{C}$  to  $+125^\circ\text{C}$ .

### PIN CONFIGURATION



### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

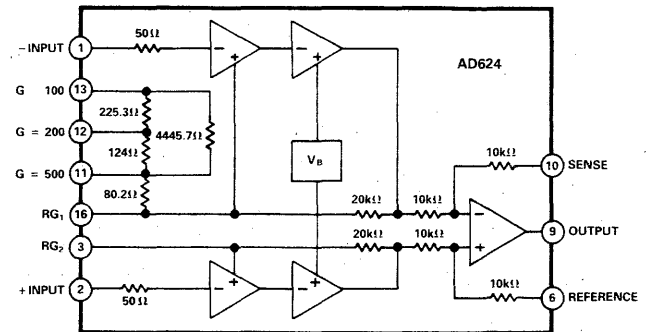


### 16-PIN CERAMIC DIP PACKAGE

### FEATURES

- Low Noise:  $0.2\mu\text{V}$  p-p 0.1Hz to 10Hz
- Low Gain TC: 10ppm
- Low Nonlinearity: 0.001% max ( $G = 1$  to 500)
- High CMRR: 130dB max ( $G = 500$ )
- Low Offset Voltage:  $25\mu\text{V}$ , max
- Low Offset Voltage Drift:  $0.25\mu\text{V}/^\circ\text{C}$  max
- Gain Bandwidth Product: 25MHz
- Pin Programmable Gains of 1, 100, 200, 500, 1000
- No External Components Required
- Internally Compensated

### AD624 FUNCTIONAL BLOCK DIAGRAM



### PRODUCT DESCRIPTION

The AD624 is a high precision ultra low noise instrumentation amplifier designed primarily for use with low level transducers, including load cells, strain gauges and pressure transducers. An outstanding combination of low noise, high gain accuracy, low gain temperature coefficient and high linearity make the AD624 ideal for use in high resolution data acquisition systems.

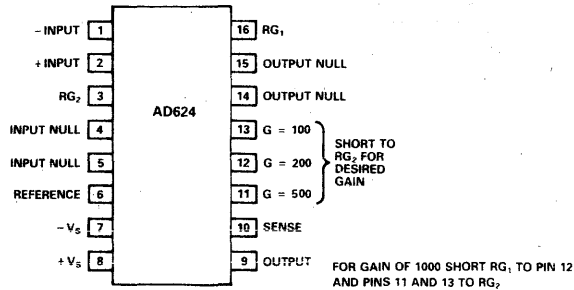
The AD624C has an input offset voltage drift of less than  $0.25\mu\text{V}/^\circ\text{C}$ , output offset voltage drift of less than  $10\mu\text{V}/^\circ\text{C}$ , CMR above 90dB at unity gain (130dB at  $G = 500$ ) and a maximum nonlinearity of 0.001% at  $G = 1$ . In addition to these outstanding dc specifications the AD624 exhibits superior ac performance as well. A 25MHz gain bandwidth product,  $5\text{V}/\mu\text{s}$  slew rate and  $15\mu\text{s}$  settling time permit the use of the AD624 in high speed data acquisition applications.

The AD624 does not need any external components for pre-trimmed gains of 1, 100, 200, 500 and 1000. Additional gains such as 250 and 330 can be programmed within a few percent accuracy with external jumpers. A single external resistor can also be used to set the 624's gain to any value in the range of 1 to 1000.

### PRODUCT HIGHLIGHTS

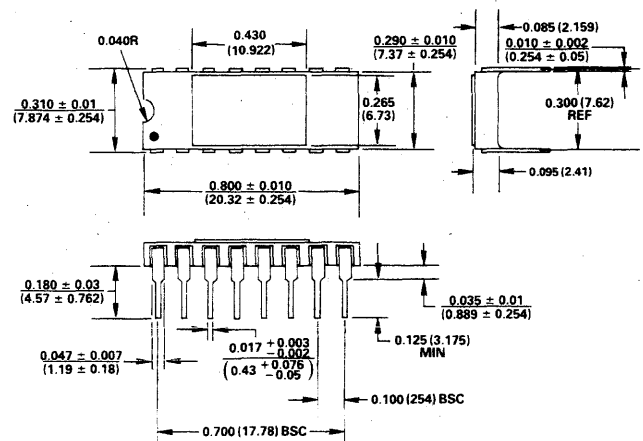
1. The AD624 offers outstanding noise performance. Input noise is typically less than  $4\text{nV}/\sqrt{\text{Hz}}$  at 1kHz.
2. The AD624 is a functionally complete instrumentation amplifier. Pin programmable gains of 1, 100, 200, 500 and 1000 are provided on the chip. Other gains are achieved through the use of a single external resistor.
3. The offset voltage, offset voltage drift, gain accuracy and gain temperature coefficients are guaranteed for all pre-trimmed gains.
4. The AD624 provides totally independent input and output offset nulling terminals for high precision applications. This minimizes the effect of offset voltage in gain ranging applications.
5. A sense terminal is provided to enable the user to minimize the errors induced through long leads. A reference terminal is also provided to permit level shifting at the output.

### PIN CONFIGURATION



### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



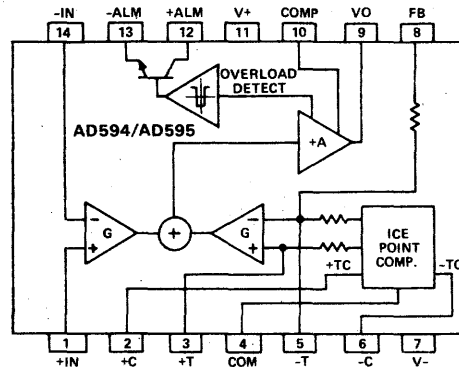
### 16-PIN CERAMIC DIP PACKAGE

## AD594\*/AD595\*

### FEATURES

- Pretrimmed for Type J (AD594) or Type K (AD595) Thermocouples
- Can Be Used with Type T Thermocouple Inputs
- Low Impedance Voltage Output: 10mV/°C
- Built-In Ice Point Compensation
- Wide Power Supply Range: +5V to ±15V
- Low Power: <1mW typical
- Thermocouple Failure Alarm
- Laser Wafer Trimmed to 1°C Calibration Accuracy
- Set-Point Mode Operation
- Self-Contained Celsius Thermometer Operation
- High Impedance Differential Input

### AD594/AD595 FUNCTIONAL BLOCK DIAGRAM

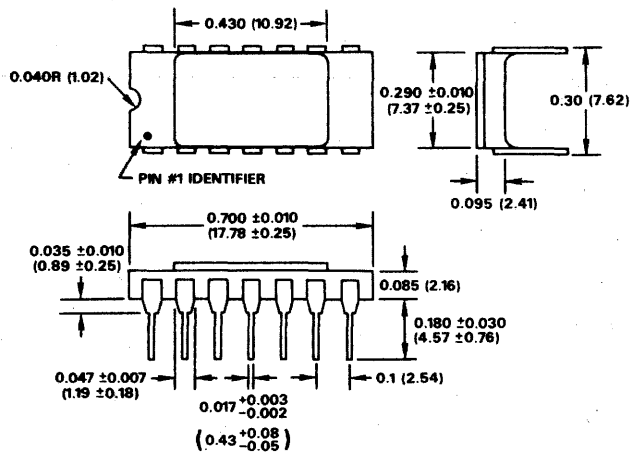


### PRODUCT HIGHLIGHTS

1. The AD594/AD595 provides cold junction compensation, amplification, and an output buffer in a single IC package.
2. Compensation, zero, and scale factor are all precalibrated by laser wafer trimming (LWT) of each IC chip.
3. Flexible pin-out provides for operation as a set-point controller or a stand-alone temperature transducer calibrated in degrees Celsius.
4. Operation at remote application sites is facilitated by low quiescent current and a wide supply voltage range of +5V to dual supplies spanning 30V.
5. Differential input rejects common-mode noise voltage on the thermocouple leads.

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



TO-116

### PRODUCT DESCRIPTION

The AD594/AD595 is a complete instrumentation amplifier and thermocouple cold junction compensator on a monolithic chip. It combines an ice point reference with a precalibrated amplifier to produce a high level (10mV/°C) output directly from a thermocouple signal. Pin-strapping options allow it to be used as a linear amplifier-compensator or as a switched output set-point controller using either fixed or remote set-point control. It can be used to amplify its compensation voltage directly, thereby converting it to a stand-alone Celsius transducer with a low-impedance voltage output.

The AD594/AD595 includes a Thermocouple Failure Alarm that indicates if one or both thermocouple leads become open. The alarm output has a flexible format which includes TTL drive capability.

The AD594/AD595 can be powered from a single ended supply (including +5V) and by including a negative supply, temperatures below 0°C can be measured. To minimize self-heating, an unloaded AD594/AD595 will typically operate with a total supply current of 160µA, but is also capable of delivering in excess of ±5mA to a load.

The AD594 is precalibrated by laser wafer trimming to match the characteristic of type J (iron-constantan) thermocouples and the AD595 is laser trimmed for type K (chromel-alumel) inputs. The temperature transducer voltages and gain control resistors are available at the package pins so that the circuit can be recalibrated for other thermocouple types by the addition of two or three resistors. These terminals also allow more precise calibration for both thermocouple and thermometer applications.

The AD594/AD595 is available in two performance grades. The C and the A versions have calibration accuracies of ±1°C and ±3°C, respectively. Both are designed to be used from 0 to +50°C, and are available in a 14-pin, hermetically sealed, side-brazed ceramic DIP.

\*Protected by U.S. Patent No. 4,029,974.

LINEAR Analog Devices

## AD9685/AD9687

### FEATURES

- 2.2ns Propagation Delay – AD9685BD/BH
- 2.7ns Propagation Delay – AD9687BD
- 0.5ns Latch Set-Up Time
- Pin-Compatible to Am685/687 but FASTER
- +5V, -5.2V Supply Voltages

### APPLICATIONS

- Ultra-High-Speed A/D Converters
- Ultra-High-Speed Line Receivers
- Peak Detectors
- Threshold Detectors

### GENERAL DESCRIPTION

The AD9685BD/BH and AD9687BD are ultra-fast comparators manufactured with a high performance bipolar process which makes it possible to obtain incredibly short propagation delays and latch set-up times.

The AD9685BD/BH is a single comparator which is pin-compatible with the Am685, but has speed capabilities that far outstrip the earlier unit. The AD9687BD is pin-for-pin compatible with the Am687 and, like its predecessor, is a dual comparator; its speed capabilities are far superior to the Am687.

Both Analog Devices units have differential inputs and complementary outputs fully compatible with ECL logic levels. Their output current levels are capable of driving 50Ω terminated transmission lines, and their high resolution make them ideally suited for a variety of analog-to-digital signal processing applications.

### AD9685BD/BH Single Comparator

A latch function allows the AD9685BD/BH to be operated in a sample-hold mode. When the Latch Enable (LE) is ECL HIGH, the comparator functions normally. When the Latch Enable is driven LOW, its outputs are locked in the logic state dictated by the input conditions at the time of the latch input transition. If the latch function is not used, the Latch Enable input should be connected to ground.

In addition to its speed advantages over the earlier Am685, the AD9685BD/BH also dissipates less power because it operates on a positive 5 volt supply instead of the 6 volts required by the AMD device.

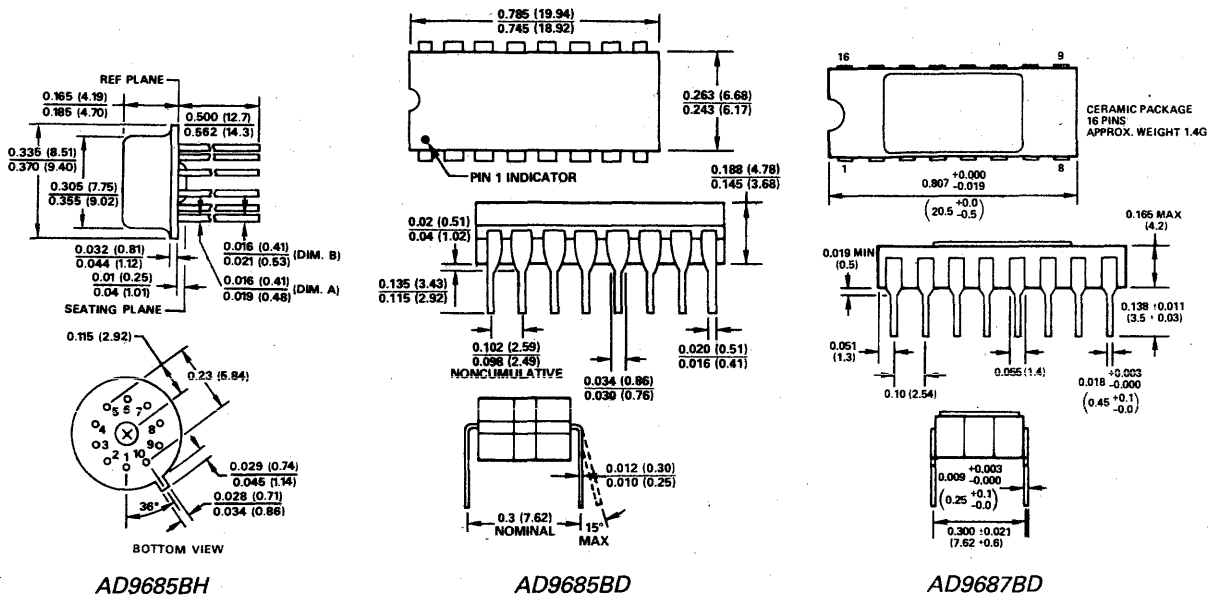
### AD9687BD Dual Comparator

The latch function of the AD9687BD provides an ability to operate the unit in either a track-hold or sample-hold mode. The latch function inputs are separated on the two comparators and are designed to be driven from the complementary outputs of a standard ECL logic gate. When LE is High and  $\overline{LE}$  is LOW, the normal comparator function is in operation. When LE is forced LOW and  $\overline{LE}$  is driven HIGH, the outputs of the comparator being exercised are locked in their existing logical states, as determined by the input conditions present at the time of arrival of the latch signal. If the latch function is not used on either one of the two comparators in the AD9687BD, the appropriate Latch Enable input should be connected to ground; the companion Latch Enable input can be left open.

The AD9687BD is basically two AD9685BD/BH units in a single package and operates in a similar fashion to a pair of the single comparators.

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



## INSTRUMENTATION AMPLIFIERS

SPECIFICATIONS (typical @  $V_S = \pm 15V$ ,  $R_L = 2k\Omega$  and  $T_A = 25^\circ C$  unless otherwise noted)

Model	Monolithic IC AD521J(K)(L)(S) <sup>1</sup>	Hybrid IC AD522A(B)(S) <sup>1</sup>	Monolithic AD524A(B)(C)(S)	Monolithic AD624A(B)(C)(S)
<b>GAIN</b>				
Range V/V	0.1 to 1000	1 to 1000	1 to 1000	1 to 1000
Nonlinearity (G = 100) - % max	0.2(0.2)(0.1)(0.2)	0.01(0.005)(0.005)	0.01(0.005)(0.003)(0.005)	0.01(0.005)(0.003)(0.005)
<b>RATED OUTPUT - V dc/ma</b>	$\pm 10/\pm 10$	$\pm 10/\pm 5$	$\pm 10/\pm 5$	$\pm 10/\pm 5$
<b>DYNAMIC RESPONSE</b>				
Small Signal (-3dB) G = 1000	40kHz	300Hz	25kHz	25kHz
Full Power Frequency	100kHz	1.5kHz	-	-
Slew Rate - V/ $\mu$ s	10	0.1	5	5
<b>OFFSET VOLTAGE</b>				
Input Offset Voltage vs. Temperature	3(1.5)(1)(1.5) $\mu$ V max 15(15)(2)(5) $\mu$ V/ $^\circ$ C	$\pm 400(200)(200)\mu$ V max $\pm 6(2)(6)\mu$ V/ $^\circ$ C	250(100)(50)(100) $\mu$ V 2(0.75)(0.5)(2) $\mu$ V/ $^\circ$ C	100(50)(25)(100) $\mu$ V 1(0.5)(0.2)(1) $\mu$ V/ $^\circ$ C
Output Offset Voltage vs. Temperature	400(1200)(100)(200) $\mu$ V max 400(150)(75)(150) $\mu$ V/ $^\circ$ C		5(3)(2)(3)mV 100(50)(25)(50) $\mu$ V/ $^\circ$ C	5(3)(2)(3)mV 100(50)(25)(50) $\mu$ V/ $^\circ$ C
<b>INPUT BIAS CURRENT - nA max</b>	$\pm 80(40)(40)(40)$	$\pm 25(15)(25)$	50(25)(15)(25)	100(50)(30)(100)
<b>INPUT IMPEDANCE</b>				
Common Mode - $\Omega$	$6 \times 10^{10}$	$10^9$	$10^9/10pF$	$10^9/10pF$
<b>COMMON MODE REJECTION RATIO</b> min @ 1k $\Omega$ Source Unbalance, CMV = $\pm 10V$				
G = 1 - dB	70(74)(74)(74) <sup>2</sup>	75(80)(85) dc to 30Hz	70(75)(80)(70)	
G = 10 - dB	90(94)(94)(94) <sup>2</sup>	90(95)(90) dc to 10Hz	90(95)(100)(90)	
G = 1000 - dB	100(110)(110)(110) <sup>2</sup>	100(110)(100) dc to 1Hz	100(115)(120)(110)	100(110)(115)(100)

**NOTES**

<sup>1</sup> Processing to MIL-STD-883B available.  
<sup>2</sup> DC to 60Hz.

## COMPARATORS

The AD9685BD/BH and AD9687BD are ultra-fast comparators which make it possible to obtain incredibly short propagation delays and latch setup times. The AD9685BD/BH is a single comparator, with the BD and BH suffixes indicating a DIP or TO-100 configuration; the AD9687BD is a dual comparator in a DIP.

The Analog Devices units are pin-for-pin compatible with the Am685 and Am687 comparators, but are far superior in speed to their forerunners. The ADI devices are an alternate source, not a second source, for the earlier units because of their greater speed capabilities but can be used as a higher-performance replacement for most applications.

SPECIFICATIONS (typical @  $+25^\circ C$  unless otherwise noted)

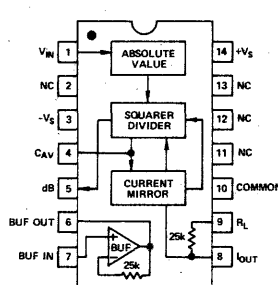
Model Number	Input Offset Voltage	Max. Input Offset Current	Input Bias Current	Input Voltage Range	Output Levels (Volts)		Propagation Delays (ns)		Latch Enable Times (nanoseconds)	
					High	Low	Input to Output High	Input to Output Low	Setup	Hold
AD9685	5mV	5 $\mu$ A	10 $\mu$ A	$\pm 5V$	-0.88	-1.75	2.2	2.2	0.5	1.0
AD9687	5mV	5 $\mu$ A	10 $\mu$ A	$\pm 5V$	-0.88	-1.75	2.7	2.7	0.5	1.0

LINEAR  
Analog Devices

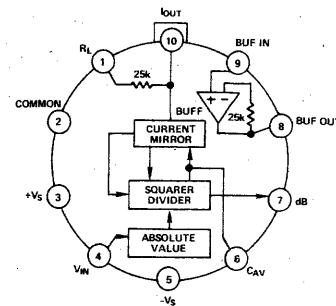
### FEATURES

- True rms-to-dc Conversion
- 200mV Full Scale
- Laser-Trimmed to High Accuracy
  - 0.5% max Error (AD636K)
  - 1.0% max Error (AD636J)
- Wide Response Capability:
  - Computes rms of ac and dc Signals
  - 1MHz -3dB Bandwidth:  $V_{rms} > 100mV$
  - Signal Crest Factor of 6 for 0.5% Error
- dB Output with 50dB Range
- Low Power: 800 $\mu$ A Quiescent Current
- Single or Dual Supply Operation
- Monolithic Integrated Circuit
- Low Cost

### AD636 FUNCTIONAL BLOCK DIAGRAMS



**TO-116  
TOP VIEW**



**TO-100  
TOP VIEW**

### PRODUCT DESCRIPTION

The AD636 is a low power monolithic IC which performs true rms-to-dc conversion on low level signals. It offers performance which is comparable or superior to that of hybrid and modular converters costing much more. Similar in operation to the AD536A, the AD636 is specified for a signal range of 0 to 200 millivolts rms. Crest factors up to 6 can be accommodated with less than 0.5% additional error, allowing accurate measurement of complex input waveforms.

The low power supply current requirement of the AD636, typically 800 $\mu$ A, allows it to be used in battery-powered portable instruments. A wide range of power supplies can be used, from  $\pm 2.5V$  to  $\pm 12V$  or a single  $+5V$  to  $+24V$  supply. The input and output terminals are fully protected; the input signal can exceed the power supply with no damage to the device (allowing the presence of input signals in the absence of supply voltage) and the output buffer amplifier is short-circuit protected.

The AD636 includes an auxiliary dB output. This signal is derived from an internal circuit point which represents the logarithm of the rms output. The 0dB reference level is set by an externally supplied current and can be selected by the user to correspond to any input level from 0dBm (774.6mV) to -20dBm (77.46mV). Frequency response ranges from 1.2MHz at a 0dBm level to over 10kHz at -50dBm.

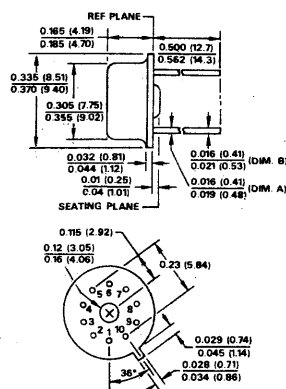
The AD636 is designed for ease of use. The device is factory-trimmed at the wafer level for input and output offset, positive and negative waveform symmetry (dc reversal), and full scale accuracy. Thus no external trims are required to achieve full rated accuracy.

The AD636 is available in two accuracy grades; the AD636J has a total error of  $\pm 0.5mV \pm 1.0\%$  of reading, and the AD636K is accurate within  $\pm 0.2mV \pm 0.5\%$  of reading. Both versions are specified for the 0 to 70 $^{\circ}C$  temperature range, and are offered in either a hermetically sealed 14-pin DIP or a 10 pin TO-100 metal can.

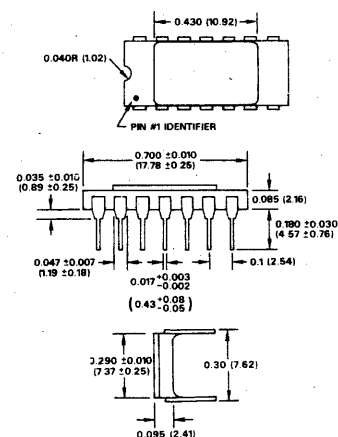
### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

#### "H" PACKAGE (TO-100)



#### "D" PACKAGE (TO-116)



### FEATURES

#### High Accuracy

- 0.02% Max Nonlinearity, 0 to 2V rms Input
- 0.10% Max Error to Crest Factor of 3

#### Wide Bandwidth

- 8MHz at 2V rms Input
- 600kHz at 100mV rms

#### Computes:

- True rms
- Square
- Mean Square
- Absolute Value

#### dB Output (-60dB Range)

#### Chip Select-Power Down Feature Allows:

- Analog "3-State" Operation
- Quiescent Current Reduction from 2.2mA to 350µA

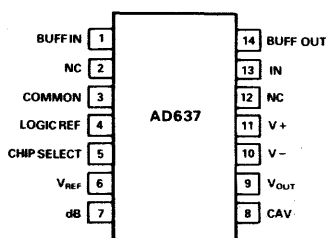
### PRODUCT DESCRIPTION

The AD637 is a complete high accuracy monolithic rms to dc converter that computes the true rms value of any complex waveform. It offers performance that is unprecedented in integrated circuit rms to dc converters and comparable to discrete and modular techniques in accuracy, bandwidth and dynamic range. A crest factor compensation scheme in the AD637 permits measurements of signals with crest factors of up to 10 with less than 1% additional error. The circuit's wide bandwidth permits the measurement of signal up to 600kHz with inputs of 200mV rms and up to 8MHz when the input levels are above 2V rms.

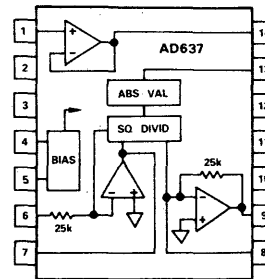
As with previous monolithic rms converters from Analog Devices, the AD637 has an auxiliary dB output available to the user. The logarithm of the rms output signal is brought out to a separate pin allowing direct dB measurement with a useful range of 60dB. An externally programmed reference current allows the user to select the 0dB reference voltage to correspond to any level between 0.1V and 2.0V rms.

A chip select connection on the AD637 permits the user to decrease the supply current from 2.2mA to 350µA during periods when the rms function is not in use. This feature facilitates the addition of precision rms measurement to remote or hand-held applications where minimum power consumption is critical. In addition when the AD637 is powered down the output goes to a high impedance state. This allows several AD637s to be tied together to form a wide-band true rms multiplexer.

### PIN CONFIGURATION



### AD637 FUNCTIONAL BLOCK DIAGRAM



The input circuitry of the AD637 is protected from overload voltages that are in excess of the supply levels. The inputs will not be damaged by input signals if the supply voltages are lost.

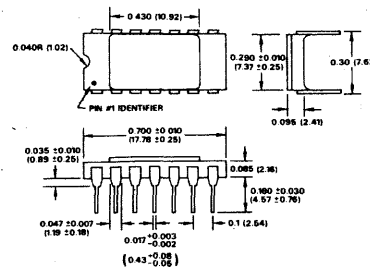
The AD637 is available in two accuracy grades (J, K) for commercial (0 to +70°C) temperature range applications and one (S) rated over the -55°C to +125°C military temperature range. All versions are available in ceramic 14-lead DIP packages.

### PRODUCT HIGHLIGHTS

1. The AD637 computes the true root-mean-square, mean square, or absolute value of any complex ac (or ac plus dc) input waveform and gives an equivalent dc output voltage. The true rms value of a waveform is more useful than an average rectified signal since it relates directly to the power of the signal. The rms value of a statistical signal is also related to the standard deviation of the signal.
2. The AD637 is laser wafer trimmed to achieve rated performance without external trimming. The only external component required is a capacitor which sets the averaging time period. The value of this capacitor also determines low frequency accuracy, ripple level and settling time.
3. The chip select feature of the AD637 permits the user to power down the device during periods of nonuse, thereby, decreasing battery drain in remote or hand-held applications.
4. The on-chip buffer amplifier can be used as either an input buffer or in an active filter configuration. The filter can be used to reduce the amount of ac ripple, thereby, increasing the accuracy of the measurement.

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).



### 14-PIN CERAMIC DIP

LINEAR  
Analog Devices

### FEATURES

- Recovers Signal from +100dB Noise
- 2MHz Channel Bandwidth
- 45V/ $\mu$ s Slew Rate
- 120dB Crosstalk @ 1kHz
- Pin Programmable Closed Loop Gains of  $\pm 1$  and  $\pm 2$
- 0.05% Closed Loop Gain Accuracy and Match
- 100 $\mu$ V Channel Offset Voltage (AD630BD)
- 350kHz Full Power Bandwidth

### PRODUCT DESCRIPTION

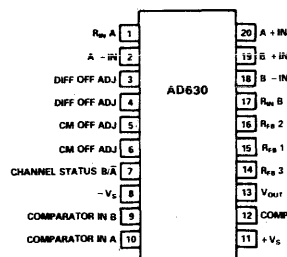
The AD630 is a high precision balanced modulator which combines a flexible commutating architecture with the accuracy and temperature stability afforded by laser wafer trimmed thin film resistors. Its signal processing applications include balanced modulation and demodulation, synchronous detection, phase detection, quadrature detection, phase sensitive detection, lock-in amplification and square wave multiplication. A network of on-board applications resistors provides precision closed loop gains of  $\pm 1$  and  $\pm 2$  with 0.05% accuracy (AD630B). These resistors may also be used to accurately configure multiplexer gains of +1, +2, +3 or +4. Alternatively, external feedback may be employed allowing the designer to implement his own high gain or complex switched feedback topologies.

The AD630 may be thought of as a precision op amp with two independent differential input stages and a precision comparator which is used to select the active front end. The rapid response time of this comparator coupled with the high slew rate and fast settling of the linear amplifiers minimize switching distortion. In addition, the AD630 has extremely low crosstalk between channels of -100dB @ 10kHz.

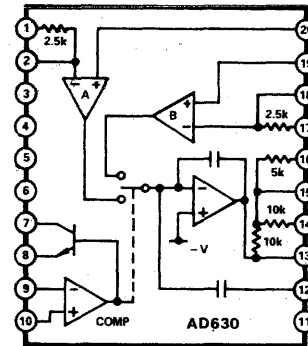
The AD630 is intended for use in precision signal processing and instrumentation applications requiring wide dynamic range. When used as a synchronous demodulator in a lock-in amplifier configuration, it can recover a small signal from 100dB of interfering noise (see lock-in amplifier application). Although optimized for operation up to 1kHz, the circuit is useful at frequencies up to several hundred kilohertz.

Other features of the AD630 include pin programmable frequency compensation, optional input bias current compensation resistors, common mode and differential offset voltage adjustment, and a channel status output which indicates which of the two differential inputs is active.

### PIN CONFIGURATION



### AD630 FUNCTIONAL BLOCK DIAGRAM

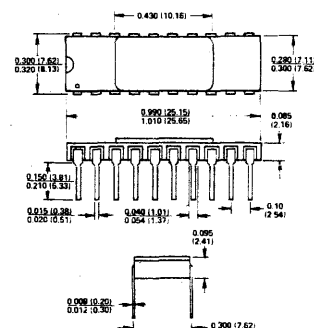


### PRODUCT HIGHLIGHTS

1. The configuration of the AD630 makes it ideal for signal processing applications such as: balanced modulation and demodulation, lock-in amplification, phase detection, and square wave multiplication.
2. The application flexibility of the AD630 makes it the best choice for many applications requiring precisely fixed gain, switched gain, multiplexing, integrating-switching functions, and high-speed precision amplification.
3. The 100dB dynamic range of the AD630 exceeds that of any hybrid or IC balanced modulator/demodulator and is comparable to that of costly signal processing instruments.
4. The op-amp format of the AD630 ensures easy implementation of high gain or complex switched feedback functions. The application resistors facilitate the implementation of most common applications with no additional parts.
5. The AD630 can be used as a two channel multiplexer with gains of +1, +2, +3 or +4. The channel separation of 100dB @ 10kHz approaches the limit which is achievable with an empty IC package.
6. The AD630 has pin-strappable frequency compensation (no external capacitor required) for stable operation at unity gain without sacrificing dynamic performance at higher gains.
7. Laser trimming of comparator and amplifying channel offsets eliminates the need for external nulling in most cases.

### OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).





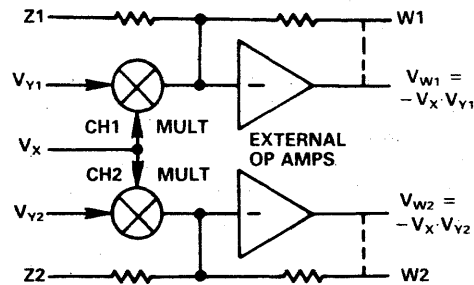
**FEATURES**

- Two Quadrant Multiplication/Division
- Two Independent Signal Channels
- Signal Bandwidth of 60MHz (I<sub>OUT</sub>)
- Linear Control-Bandwidth of 5MHz
- Fully-Calibrated, Monolithic Circuit

**APPLICATIONS**

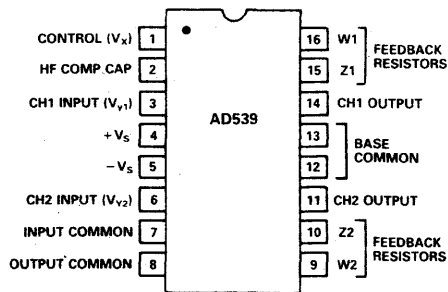
- Precise AGC and VCA Systems
- Voltage-Controlled Filters
- Video-Signal Processing
- High-Speed Analog Division
- Automatic Signal-Leveling
- Square-Law Gain/Loss Control

**AD539 FUNCTIONAL BLOCK DIAGRAM**

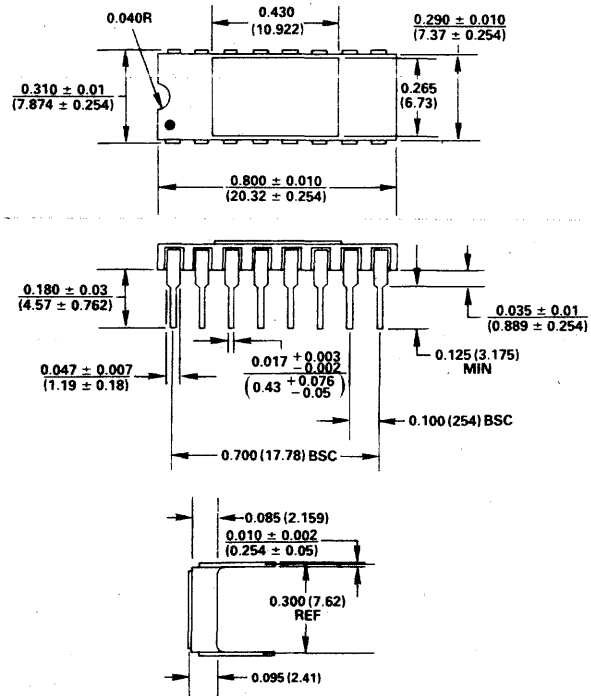


**PIN CONFIGURATION & DIMENSIONS**

Dimensions shown in inches and (mm).



**TO-116**



**PRODUCT DESCRIPTION**

The AD539 is a low-distortion analog multiplier having two identical signal channels (Y1 and Y2), with a common X-input providing linear control of gain. Excellent ac characteristics up to video frequencies and a 3dB bandwidth of over 60MHz are provided. Although intended primarily for applications where speed is important the circuit exhibits good static accuracy in "computational" applications. Scaling is accurately determined by a band-gap voltage reference and all critical parameters are laser-trimmed during manufacture.

The full bandwidth can be realized over most of the gain range using the AD539 with simple resistive loads of up to 100Ω. Output voltage is restricted to a few hundred millivolts under these conditions. Using external op amps in conjunction with the on-chip scaling resistors, accurate multiplication and large output voltages can be achieved, but with a reduction in bandwidth typically to 25MHz.

The two channels provide flexibility. In single-channel applications they may be used in parallel, to double the output current, or in series, to achieve a square-law gain function with a control range of over 100dB, or differentially, to reduce distortion. Alternatively, they may be used independently, as in audio stereo applications, with low crosstalk between channels. Voltage-controlled filters and oscillators using the "state-variable" approach are easily designed, taking advantage of the dual channels and common control. The AD539 can also be configured as a divider with signal bandwidths up to 15MHz.

Power consumption is only 135mW using the recommended ±5V supplies. The AD539 is available in three versions: the "J" and "K" grades are specified for 0 to +70°C operation and "S" grade is guaranteed over the extended range of -55°C to +125°C. All versions are packaged in 16-pin DIPs.

**COMPUTATIONAL CIRCUITS**
**SPECIFICATIONS (typical @ +25°C and rated supply voltage unless noted otherwise)**

Model	Transfer Function	Full Scale Accuracy - % Max	Accuracy vs. Temperature %/°C	X Nonlinearity % of Full Scale	Y Nonlinearity % of Full Scale	Bandwidth Small Signal MHz	Operating Power Supply Volts	Operating Temperature Range <sup>1</sup>
AD531J	$XY/kl_z$	2	0.04	0.8	0.3	1	±15 to ±18	C
AD531K	$XY/kl_z$	1	0.03	0.5	0.2	1	±15 to ±18	C
AD531L	$XY/kl_z$	0.5	0.01	0.3	0.2	1	±15 to ±18	C
AD531S	$XY/kl_z$	1	0.03 max	0.5	0.2	1		M
AD532J	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10}$	2	0.04	0.8	0.3	1	±15 to ±18	C
AD532K	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10}$	1	0.03	0.5	0.2	1	±15 to ±18	C
AD532S	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10}$	1	0.04 max	0.5	0.2	1		M
AD533J	$XY/10$	2	0.04	0.8	0.3	1	±15 to ±18	C
AD533K	$XY/10$	1	0.03	0.5	0.2	1	±15 to ±18	C
AD533L	$XY/10$	0.5	0.01	0.5	0.2	1	±15 to ±18	C
AD533S	$XY/10$	1	0.01	0.5	0.2	1		M
AD534J	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10 + Z_2}$	1	0.022	0.4	0.01	1	±8 to ±18	C
AD534K	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10 + Z_2}$	0.5	0.015	0.3 max	0.01 max	1	±8 to ±18	C
AD534L	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10 + Z_2}$	0.25	0.008	0.12 max	0.01 max	1	±8 to ±18	C
AD534S	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10 + Z_2}$	1	0.02 max	0.4	0.01	1	±8 to ±18	M
AD534T	$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10 + Z_2}$	0.5	0.01 max	0.3 max	0.01 max	1	±8 to ±18	M
AD539J	$V_x V_{y1}, V_x V_{y2}$ (Note 2)	2.5	5% Total <sup>3</sup>	0.5dB <sup>4</sup>	—	30 min	±4.5 to ±16.5	C
AD539K	$V_x V_{y1}, V_x V_{y2}$ (Note 2)	1.5	2% Total <sup>3</sup>	0.2dB <sup>4</sup>	—	30 min	±4.5 to ±16.5	C
AD539S	$V_x V_{y1}, V_x V_{y2}$ (Note 2)	2.5	5% Total <sup>3</sup>	0.5dB <sup>4</sup>	—	30 min	±4.5 to ±16.5	M

**NOTES**
<sup>1</sup>C: 0 to +70°C, M: -55°C to +125°C

<sup>2</sup>Channels 1 & 2

<sup>3</sup>Multiplier scaling voltage

<sup>4</sup>Absolute gain error

**RMS-TO-DC CONVERTERS**
**SPECIFICATIONS (typical @ +25°C and rated supply voltage unless otherwise noted)**

Model	Transfer Function	Conversion Accuracy	Conversion Accuracy vs. Temperature	Input Signal Range	I <sub>OUT</sub> Scale Factor	Bandwidth Full Scale	Operating Power Supply Volts	Operating Temperature Range <sup>1</sup>
AD536AJ	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±5mV ±0.5% of RDG max	±(0.1mV ±0.01% of RDG)/°C	±10V	40μA/V rms	2MHz	+5 to ±18	C
AD536AK	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±2mV ±0.2% of RDG max	±(0.05mV ±0.005% of RDG)/°C	±10V	40μA/V rms	2MHz	+5 to ±18	C
AD536AS	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±5mV ±0.5% of RDG max	±(0.1mV ±0.01% of RDG)/°C	±10V	40μA/V rms	2MHz	+5 to ±18	M
AD636J	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±0.5mV ±1.0% of RDG max	±(0.1mV ±0.001% of RDG)/°C	±200mV	100μA/V rms	1.3MHz	+2/-2.5 to ±12	C
AD636K	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±0.2mV ±0.5% of RDG max	±(0.1mV ±0.0005% of RDG)/°C	±200mV	100μA/V rms	1.3MHz	+2/-2.5 to ±12	C
AD637J	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±1mV ±0.2% of RDG max	±0.4%	±10V	—	5MHz	±3.0 to ±18	C
AD637K	$V_{OUT} = \sqrt{\text{avg.}(V_{IN})^2}$	±0.5mV ±0.1% of RDG max	±0.2%	±10V	—	5MHz	±3.0 to ±18	C

**NOTE**
<sup>1</sup>C: 0 to +70°C, M: -55°C to +125°C

**BALANCED MODULATOR-DEMODULATOR**
**SPECIFICATIONS (typical @ +25°C and ±15V dc unless otherwise noted)**

Model	AD630	Model	AD630
<b>Gain</b>		<b>AC Characteristics</b>	
Gain Accuracy ±1, ±2	0.05%	Comparator Hysteresis	1mV (0 min)
Gain Match	0.03%	Comparator Response Time	200ns
		Small Signal -3dB Bandwidth	2MHz
		Slew Rate	40V/μs
<b>Input Characteristics, Amplifiers &amp; Comparator</b>		<b>System Characteristics</b>	
Input Offset Voltage	250μV	CMRR	90dB
Input Bias Current	250nA	PSRR	80dB
Input Offset Current	50nA	Channel Separation @ 10kHz	-80dB
Input Impedance	15MΩ  1.5pF		

# ABBREVIATIONS OF COMPANY NAMES

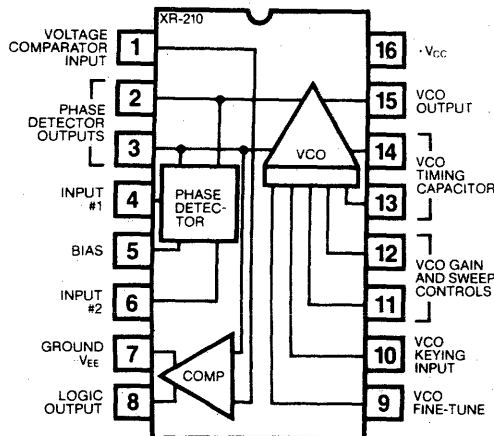
<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.			<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Harris</b>	Harris Semiconductor	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Heurikon</b>	Heurikon Corp.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Hitachi</b>	Hitachi America, Ltd.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>Holt</b>	Holt Inc.		
<b>Analogic</b>	Analogic	<b>HP</b>	Hewlett-Packard		
<b>Analog Sys</b>	Analog Systems	<b>Hughes</b>	Hughes Aircraft, Solid State Products	<b>Panasonic</b>	Panasonic
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Pico Design</b>	Pico Design
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>Polycore</b>	Polycore Electronics
<b>APM</b>	Applied Microsystems Corp.			<b>Plessey</b>	Plessey Semiconductors
<b>Appl Sys</b>	Applied Systems Corp.			<b>PMI</b>	Precision Monolithics, Inc.
<b>APT</b>	Applied Microtechnology			<b>PragDes</b>	Pragmatic Design Inc.
<b>Aptek</b>	Aptek Microsystems			<b>Pro-Log</b>	Pro-Log Corp.
<b>Array Tech</b>	Array Technology				
<b>AWI</b>	AWI Electronics	<b>ICC</b>	International Cybernetics	<b>Quay</b>	Quay Corp.
		<b>IDT</b>	Integrated Device Technology		
		<b>IMI</b>	Integrated Microcircuits, Inc.	<b>Raytheon</b>	Raytheon Semiconductor
		<b>IMP</b>	International Microelectronic Products	<b>RCA</b>	RCA Solid State Division
				<b>RCI Data</b>	RCI Data
<b>Barvon</b>	Barvon Research	<b>IMS</b>	Industrial MicroSystems Inc.	<b>RELMS</b>	Relational Memory Systems
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>Infosphere</b>	Infosphere	<b>Reticon</b>	Reticon
<b>Burr-Brown</b>	Burr-Brown	<b>Inmos</b>	Inmos	<b>RIFA</b>	RIFA
		<b>IntCirEng</b>	Integrated Circuit Engineering	<b>Rockwell</b>	Rockwell, Microelectronic Devices
		<b>IntCirSys</b>	Integrated Circuit Systems	<b>RTC</b>	Riehl Time Corporation
		<b>IntCompSys</b>	Integrated Computer Systems		
<b>CAE</b>	Computer Aided Engineering	<b>Int Tech</b>	Integrated Technology Corp.	<b>Sanyo</b>	Sanyo
<b>Cal Devices</b>	California Devices	<b>Intech</b>	Intech Microcircuits Div.	<b>SBE</b>	SBE, Inc.
<b>Cermetek</b>	Cermetek	<b>Intel</b>	Intel	<b>SEEQ</b>	SEEQ Technology, Inc.
<b>CGRS</b>	CGRS Microtech Inc.	<b>Interdesign</b>	Interdesign	<b>SPI</b>	Semi Processes Inc.
<b>Cherry</b>	Cherry Semiconductor	<b>Intersil</b>	Intersil	<b>Siemens</b>	Siemens
<b>CIC</b>	Custom Integrated Circuits	<b>Intronics</b>	Intronics	<b>Si-Fab</b>	Si-Fab
<b>CirTech</b>	Circuit Technology	<b>ITT</b>	ITT Semiconductors	<b>Signetics</b>	Signetics
<b>Citel</b>	Citel			<b>SGS</b>	SGS Semiconductor
<b>Comlinear</b>	Comlinear Corporation	<b>Kinetic Sys</b>	Kinetic Systems	<b>Sharp</b>	Sharp
<b>CMA</b>	Custom MOS Arrays	<b>Kontron</b>	Kontron Electronics	<b>Silicon G</b>	Silicon General
<b>Comark</b>	Comark Corp.			<b>Siliconix</b>	Siliconix
<b>Comdial</b>	Comdial Semiconductor			<b>Silicon Sys</b>	Silicon Systems Inc.
<b>Comp Auto</b>	Computer Automation	<b>Lambda</b>	Lambda Semiconductor	<b>Siltronics</b>	Siltronics
<b>Compas</b>	Compas Microsystems	<b>Linear Tech</b>	Linear Technology	<b>SMC</b>	Standard Microsystems Corp.
<b>Cont Logic</b>	Control Logic Inc.	<b>LSI Comp</b>	LSI Computer Systems	<b>S MOS</b>	S MOS Systems
<b>Control Sys</b>	Control Systems Microsystems Div.	<b>LSI Logic</b>	LSI Logic Corporation	<b>Solarise</b>	Solarise Enterprises
<b>CreMicro</b>	Creative Micro Systems			<b>Solitron</b>	Solitron Devices
<b>Cromemco</b>	Cromemco, Inc.	<b>Master Logic</b>	Master Logic Corporation	<b>Sprague</b>	Sprague Electric Company
<b>Cubit</b>	Cubit Inc.	<b>Matrix</b>	Matrix Corp.	<b>SSM</b>	Solid State Micro Technology for Music
<b>Curtis</b>	Curtis Electro Devices, Inc.	<b>Matrox</b>	Matrox Electronic Systems	<b>SSS</b>	Solid State Scientific
<b>Cybernetic</b>	Cybernetic Micro Systems	<b>Maxim</b>	Maxim Integrated Products	<b>Stag</b>	Stag Microsystems
<b>Cybersys</b>	Cybersystems	<b>MCC</b>	Micro-Computer Control	<b>STC</b>	Storage Technology Corp.
<b>Cybertek</b>	Cybertek Inc.	<b>MCE</b>	MCE Electronics	<b>STD</b>	STD Microsystems
		<b>Micrel</b>	Micrel	<b>Struc Des</b>	Structured Design Inc.
		<b>Micro Innov</b>	Micro Innovators	<b>Stynetic</b>	Stynetic Systems
		<b>Micropac</b>	Micropac Industries	<b>Sunrise</b>	Sunrise Electronics
<b>Data General</b>	Data General	<b>Micro Net</b>	Micro Networks	<b>Sunshine</b>	Sunshine Semiconductor
<b>Data I/O</b>	Data I/O	<b>Micro Pwr</b>	Micro Power Systems	<b>Supertex</b>	Supertex Inc.
<b>Data Trans</b>	Data Translation	<b>Micro Sci</b>	Micro Sciences Corp.	<b>Symtek</b>	Symtek Corp.
<b>Datel</b>	Datel	<b>Micro Tech</b>	Microcircuits Technology	<b>Synartek</b>	Synartek
<b>Datricon</b>	Datricon Corporation	<b>Micro-Link</b>	Micro-Link Corporation	<b>Sys Innov</b>	Systems Innovations
<b>DDC</b>	Data Devices Corporation	<b>Micron</b>	Micron Technology		
<b>DEC</b>	Digital Equipment Corporation	<b>MillerTron</b>	MillerTronics	<b>Tau Zero</b>	Tau Zero Inc.
<b>Die-Tech</b>	Die-Tech	<b>Miller</b>	Miller Technology	<b>Technitrol</b>	Technitrol
<b>Digelec</b>	Digelec Corp.	<b>Mitel</b>	Mitel Semiconductor	<b>Tektronix</b>	Tektronix
<b>Digitel</b>	Digitel, Inc.	<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Teledyne C</b>	Teledyne Crystalonics
<b>Dionics</b>	Dionics Inc.	<b>MMI</b>	Monolithic Memories, Inc.	<b>Teledyne P</b>	Teledyne Philbrick
<b>Dist Comp</b>	Distributed Computer Systems			<b>Teledyne S</b>	Teledyne Semiconductor
<b>Divers Tech</b>	Diversified Technology	<b>Mostek</b>	Monolithic Systems Corp.	<b>Telefunken</b>	Telefunken
		<b>Motorola</b>	Motorola Semiconductor	<b>Telmos</b>	Telmos
		<b>MRC</b>	MRC Systems	<b>Teltone</b>	Teltone Corporation
		<b>Murray</b>	Murray Consulting	<b>TI</b>	Texas Instruments
		<b>Monosil</b>		<b>Third Domain</b>	Third Domain
<b>E-HI</b>	E-H International, Inc.			<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
<b>EDI</b>	Electronic Designs Inc.	<b>National</b>	National Semiconductor	<b>Toshiba</b>	Toshiba America
<b>Elind</b>	Elind Electronica Industriale	<b>NCM</b>	NCM Corp.	<b>Trans-Data</b>	Trans-Data
<b>EMM-SESCO</b>	EEM-SESCO	<b>NCR</b>	NCR Corp., Microelectronics Division	<b>TRW</b>	TRW LSI Products
<b>Emulogic</b>	Emulogic Inc.				
<b>ETI Micro</b>	ETI Micro	<b>Varix</b>	Varix Corp.	<b>Unitrode</b>	Unitrode
<b>Exar</b>	Exar Integrated Systems	<b>VLSI Design</b>	VLSI Design Associates	<b>Universal</b>	Universal Semiconductor, Inc.
<b>Exel</b>	Exel Microelectronics	<b>VTI</b>	VLSI Technology, Inc.		
		<b>Votrax</b>	Votrax	<b>Weitek</b>	Weitek Corporation
<b>Fairchild</b>	Fairchild			<b>Western</b>	Western Digital
<b>Ferranti</b>	Ferranti Electric	<b>Xicor</b>	Xicor, Inc.	<b>Wintek</b>	Wintek Corp.
<b>Force</b>	Force Computers	<b>Xycom</b>	Xycom		
<b>Fujitsu A</b>	Fujitsu America			<b>Zendex</b>	Zendex Corp.
<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.	<b>NEC</b>	NEC Electronics	<b>Zilog</b>	Zilog
		<b>Nitron</b>	Nitron	<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrex</b>	Zytrex Corp.

## Phase-Locked Loops

### XR-210 FSK MODULATOR/ DEMULATOR

The XR-210 is a highly versatile monolithic phase-locked loop system especially designed for data communications. It is particularly well suited for FSK modulation/demodulation (MODEM) applications, frequency synthesis, tracking filters and tone decoding. The XR-210 operates over a power supply range of 5V to 26V, and over a frequency band of 0.5 Hz to 20 MHz. The circuit can accommodate analog signals between 300  $\mu$ V and 3V, and can interface with conventional DTL, TTL, and ECL logic families.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Digital Programming Capability  
 RS-232C Compatible Demodulator Output  
 ON-OFF Keying & Sweep Capability  
 Wide Tracking Range  $\pm 1\%$  to  $\pm 50\%$   
 Good Temperature Stability 200 ppm/ $^{\circ}$ C  
 High Current Logic Output 50 mA  
 Independent "Mark" and "Space" Frequency Adjustment  
 VCO Duty Cycle Control

#### APPLICATIONS

Data Synchronization	Signal Conditioning
FSK Generation	Tone Decoding
Frequency Synthesis	FSK Demodulation
Tracking Filter	FM Detection
FM and Sweep Generation	Wideband Discrimination

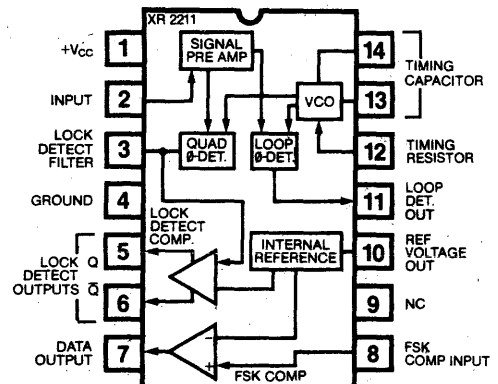
#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-210M	Ceramic	-55 $^{\circ}$ C to +125 $^{\circ}$ C
XR-210CN	Ceramic	0 $^{\circ}$ C to +70 $^{\circ}$ C

### XR-2211 FSK DEMODULATOR/ TONE DECODER

The XR-2211 is a monolithic phase-locked loop system especially designed for data communications, and particularly well-suited for FSK modem applications. It has a supply voltage range of 4.5V to 20V and a wide frequency range of 0.01 Hz to 300 kHz. The circuit accommodates analog signals between 2 mV and 3V, and interfaces with conventional DTL, TTL, and ECL logic families. The XR-2211 consists of a basic PLL for tracking an input signal within the passband, a quadrature phase detector for carrier detection, and an FSK voltage comparator for FSK demodulation. Independent external components set the center frequency, bandwidth, and output delay.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Wide Frequency Range 0.01 Hz to 300 kHz  
 Wide Supply Voltage Range 4.5V to 20V  
 DTL/TTL/ECL Logic Compatibility  
 Wide Dynamic Range 2 mV to 3V rms  
 Adjustable Tracking Range  $\pm 1\%$  to  $\pm 80\%$   
 Excellent Temperature Stability 20 ppm/ $^{\circ}$ C, Typical  
 FSK Demodulation with Carrier-Detection

#### APPLICATIONS

FSK Demodulation	FM Detection
Data Synchronization	Carrier Detection
Tone Decoding	

#### ORDERING INFORMATION

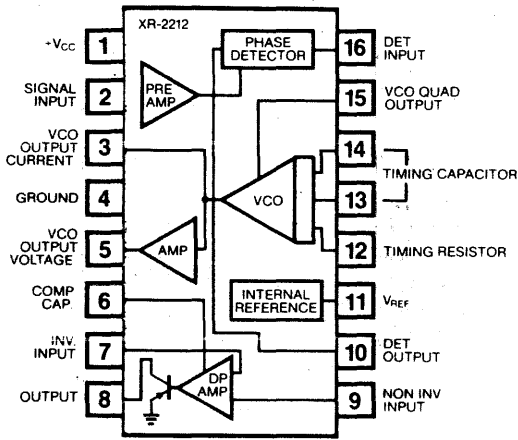
Part Number	Package	Operating Temperature
XR-2211M	Ceramic	-55 $^{\circ}$ C to +125 $^{\circ}$ C
XR-2211N	Ceramic	-40 $^{\circ}$ C to +85 $^{\circ}$ C
XR-2211P	Plastic	-40 $^{\circ}$ C to +85 $^{\circ}$ C
XR-2211CN	Ceramic	0 $^{\circ}$ C to +70 $^{\circ}$ C
XR-2211CP	Plastic	0 $^{\circ}$ C to +70 $^{\circ}$ C

# Phase-Locked Loops

## XR-2212 PRECISION PHASE-LOCKED LOOP

The XR-2212 is an ultra-stable monolithic phase-locked loop (PLL) system especially designed for data communication and control system applications. It is ideally suited for frequency synthesis, FM detection, and tracking filter applications. The circuit consists of a PLL system made up of an input preamplifier, a phase detector, a stable voltage-controlled oscillator (VCO), and a high-gain differential amplifier. The center frequency, bandwidth, and tracking range of the PLL are controlled independently by the choice of external components.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Quadrature VCO Outputs
- Wide Frequency Range: 0.01 Hz to 300 kHz
- Wide Supply Voltage Range: 4.5V to 20V
- Adjustable Tracking Range:  $\pm 1\%$  to  $\pm 80\%$
- Excellent Temperature Stability: 20 ppm/ $^{\circ}\text{C}$ , Typical

### APPLICATIONS

- Frequency Synthesis
- Data Synchronization
- FM Detection
- Signal Conditioning
- FSK Demodulation
- Tracking Filters
- Clock Extraction

### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2212M	Ceramic	-55 $^{\circ}\text{C}$ to +125 $^{\circ}\text{C}$
XR-2212N	Ceramic	-40 $^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$
XR-2212P	Plastic	-40 $^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$
XR-2212CN	Ceramic	0 $^{\circ}\text{C}$ to +70 $^{\circ}\text{C}$
XR-2212CP	Plastic	0 $^{\circ}\text{C}$ to +70 $^{\circ}\text{C}$

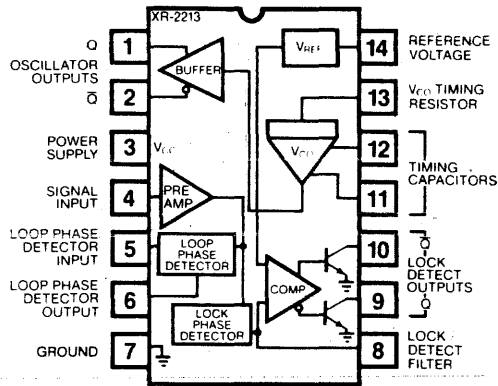
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## XR-2213 PRECISION PHASE-LOCKED LOOP/TONE DECODER

The XR-2213 is a high-stability phase-locked loop system designed for control systems and tone detection applications. The circuit features a high-stability VCO, an input preamplifier and a high-gain voltage comparator, in addition to loop and quadrature-phase detectors.

The XR-2213 has an operating frequency range of 0.01 Hz to 300 kHz. The input preamplifier allows an input signal range of 2 mV to 3 volts rms, while the VCO and comparator supply TTL-compatible output signals. The VCO and voltage comparator provide both Q and  $\bar{Q}$  outputs for greater flexibility and reduced external parts count. All of the loop parameters are independently adjustable by the choice of external components.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Wide Frequency Range: 0.01 Hz to 300 kHz
- Wide Supply Voltage Range: 4.5V to 15V
- Uncommitted VCO Q and  $\bar{Q}$  Outputs
- Wide Dynamic Input Voltage Range: 2 mV to 3V rms
- Excellent VCO Stability: 20 ppm/ $^{\circ}\text{C}$ , Typical

### APPLICATIONS

- Tone Detection
- Frequency Synthesis
- FM Detection
- Tracking Filters

### ORDERING INFORMATION

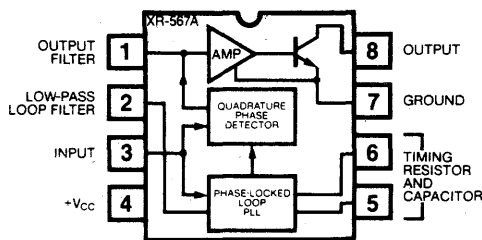
Part Number	Package	Operating Temperature
XR-2213M	Ceramic	-55 $^{\circ}\text{C}$ to +125 $^{\circ}\text{C}$
XR-2213N	Ceramic	-40 $^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$
XR-2213P	Plastic	-40 $^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$
XR-2213CN	Ceramic	0 $^{\circ}\text{C}$ to +70 $^{\circ}\text{C}$
XR-2213CP	Plastic	0 $^{\circ}\text{C}$ to +70 $^{\circ}\text{C}$

## Tone Decoders

### XR-567/567A MONOLITHIC TONE DECODER

The XR-567 is a monolithic phase-locked loop system designed for general purpose tone and frequency decoding. It offers a wide frequency band of 0.01 Hz to 500 kHz, and has a logic compatible output capable of sinking up to 100 mA of load current. Four independent external components determine the bandwidth, center frequency, and output delay. The XR-567A offers improved frequency accuracy and drift characteristics over the standard 567. The improved design enhances overall circuit performance, while reducing initial circuit adjustments.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Direct Replacement for SE/NE567  
 Adjustable Bandwidth 0% to 14%  
 Logic Compatible Output  
 with Current Sinking Capability 100 mA  
 Adjustable Center Frequency 0.01 Hz to 500 kHz  
 Inherent Immunity to False Signals  
 High Rejection of Out-of-Band Signals and Noise  
 Adjustable Range Frequency by External Resistor over 20:1 range

#### APPLICATIONS

Touch-Tone<sup>®</sup> Decoding  
 Sequential Tone Decoding  
 Communications Paging  
 Ultrasonic Remote Control and Monitoring  
 Carrier-Tone Transceiver  
 Wireless Intercom  
 Precision Oscillator

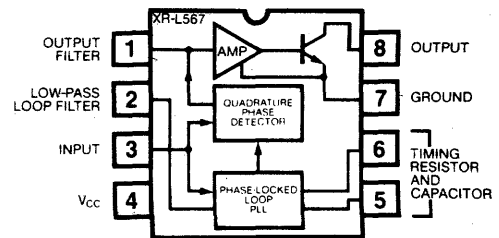
#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-567M/AM	Ceramic	-55°C to +125°C
XR-567CN/ACN	Ceramic	0°C to +70°C
XR-567CP/ACP	Plastic	0°C to +70°C

### XR-L567 MONOLITHIC TONE DECODER

The XR-L567 is a micropower phase-locked loop (PLL) circuit designed for general purpose tone and frequency decoding. It is a direct replacement for the popular 567-type tone decoder ICs, designed for applications requiring very low power dissipation. The XR-L567 offers approximately 1/10th the power dissipation of the conventional 567-type tone decoder, without sacrificing its key features, such as oscillator stability, frequency selectivity and detection threshold. At 5-volt operation, typical quiescent power dissipation is less than 4 mW.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Very Low Power Dissipation < 4 mW at 5V  
 Adjustable Bandwidth 0% to 14%  
 Logic Compatible Output with  
 10 mA Current Sinking Capability  
 Highly Stable Center Frequency  
 Adjustable Center Frequency 0.01 Hz to 50 kHz  
 Inherent Immunity to False Signals  
 High Rejection of Out-of-Band  
 Signals and Noise  
 Adjustable Frequency Range by External Resistor over 20:1 range

#### APPLICATIONS

Battery Operated Tone Detection  
 Touch-Tone<sup>®</sup> Decoding  
 Communications Paging  
 Ultrasonic Remote Control  
 Telemetry Decoding

#### ORDERING INFORMATION

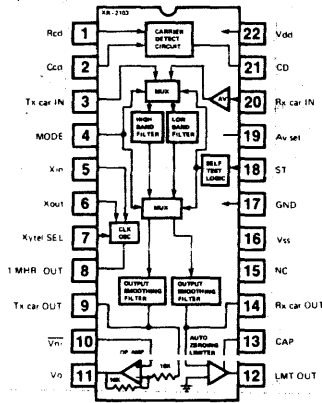
Part Number	Package	Operating Temperature
XR-L567CN	Ceramic	0°C to +70°C
XR-L567CP	Plastic	0°C to +70°C

# Filters

## XR-2103 FSK MODEM FILTER

The XR-2103 is a Monolithic Switched-Capacitor Filter designed to perform the complete filtering function necessary for a Bell 103 Compatible Modem. The XR-2103 is specifically intended for use with the XR-14412 Modulator/Demodulator to form a complete stand alone two-chip Modem. In addition to complete high and low bandpass filters, the XR-2103 contains internal mode switching, auto-zeroing limiter and dedicated duplexer op amp. An on-board carrier detect circuit is also included to complete the over-all system. Designed for crystal-controlled operation, the XR-2103 may operate from a 4.0 MHz crystal or 1 MHz input from the XR-14412. A self-test circuit is included to complement that of the XR-14412.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Single 5 Volt Operation
- 4.0 MHz/1 MHz Clock Input
- Complete On-board Output Active Filters
- Low Supply Current
- Internal Answer/Originate Mode Switching
- Programmable Input Receive Gain

### APPLICATIONS

- Bell 103 Transmit/Receive Filtering
- Complement to XR-14412 or other Modulator/Demodulators

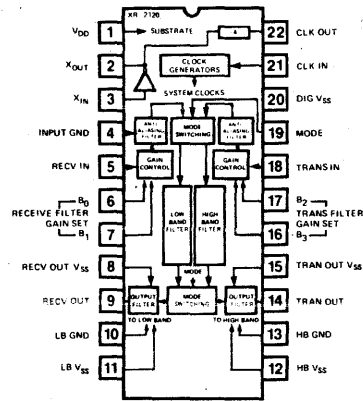
### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2103CP	Plastic	0°C to 70°C
XR-2103CN	Ceramic	0°C to 70°C

## XR-2120 PSK MODEM FILTER

The XR-2120 is a self-contained bandpass filter set designed for realization of Bell 212A compatible 1200 bits/sec PSK MODEMS. The XR-2120 utilizes CMOS technology and switched capacitor circuit techniques to minimize external components to a single crystal or frequency source. Contained in the device are two complete bandpass filters centered around the Bell standard 1200 Hz and 2400 Hz send and receive frequencies. These filters also provide compromise line equalization. Additional features included are digitally programmable transmit and receive gains as well as input anti-aliasing and complete output smoothing filters. Separate  $V_{SS}$  pins for transmit, receive, and digital sections are provided to minimize crosstalk.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- On-board Crystal Oscillator With Buffered Output
- Internal Anti-aliasing Filters
- Complete On-board Output Active Filters
- Digitally Programmable Transmit and Receive Gains
- MODE Input Internally Switches Filters for Answer/Originate
- Single or Split Supply Operation
- Center Frequencies Movable with Input Clock
- High-Impedance Inputs (100 kΩ min)
- Low Supply Current
- 1% Center Frequency Accuracy
- Separate CLK IN and CLK OUT Pins

### APPLICATIONS

- Bell 212A Transmit/Receive Filtering
- Answer Back Signal Filtering

### ORDERING INFORMATION

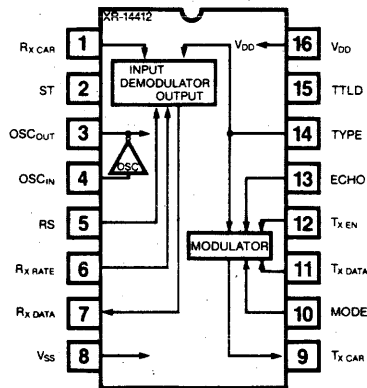
Part Number	Package	Operating Temperature
XR-2120CN	Ceramic	0°C to +70°C
XR-2120CP	Plastic	0°C to +70°C

## Modem Circuits

### XR-14412 FSK MODEM SYSTEM

The XR-14412 contains all the necessary circuitry to construct a complete FSK modulator/demodulator (MODEM) system. Included is circuitry for pin-programmable frequency bands, either U.S. or foreign (CCITT) standards for low-speed MODEMS. The XR-14412 provides TTL-compatible inputs and outputs. Included in the XR-14412 are features for self-testing and an echo suppression tone generator. The XR-14412 utilizes complementary MOS technology for low-power operation.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Simplex, Half-Duplex, and Full Duplex Operation
- Crystal Controlled
- Answer or Originate Modes
- Single Supply Operation
- Self-test Mode
- Selectable Data Rates 300 or 600 bps
- TTL, or CMOS-Compatible Inputs and Outputs
- Echo Suppressor Disable Tone Generator
- U.S. or Foreign (CCITT) Compatible

#### APPLICATIONS

- Stand-Alone MODEMS
- Remote Terminals
- Acoustical Couplers
- Built-in MODEMS

#### ORDERING INFORMATION

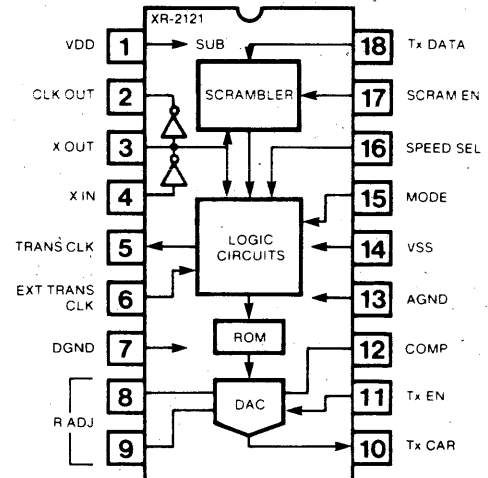
Part Number	Package	Operating Temperature
XR-14412FN	Ceramic	-40°C to +85°C
XR-14412VN	Ceramic	-40°C to +85°C
XR-14412FP	Plastic	-40°C to +85°C
XR-14412VP	Plastic	-40°C to +85°C

### XR-2121 BELL 212A MODULATOR

The XR-2121 is designed to provide the complete modulator function for the Bell Standard 212A PSK Modem. It can be used with the XR-2120 PSK Filter, XR-2122 Demodulator, and XR-2125 Sync/Async Converter to provide a four chip solution for realizing this 1200/300 bps Modem System.

The XR-2121 Modulator provides either a 1200 bps PSK or 0 to 300 bps FSK output. Crystal controlled operation offers extremely accurate and stable 1200/2400 Hz carriers for the PSK and 1170/2125 Hz carriers for the FSK. An enable/disable pin is provided for blanking the modulator output. A transmit clock of 1200 Hz is also provided for synchronization of the terminal and other facilities. The XR-2121 is designed to operate from dual 5-volt power supplies and utilizes CMOS technology for low power consumption.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Bell Standard 212A Compatible
- 6-Bit Synthesized Sine Wave Output
- Internal Scrambler
- Enable/Disable Input
- 1200 Hz Transmit Clock Output
- Crystal Controlled with Buffered Clock Output
- External Clock Input
- CMOS/TTL Compatible Inputs

#### APPLICATIONS

- Stand-Alone Modems
- Remote Terminals
- Built-In Modems

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2121CN	Ceramic	0°C to +70°C
XR-2121CP	Plastic	0°C to +70°C



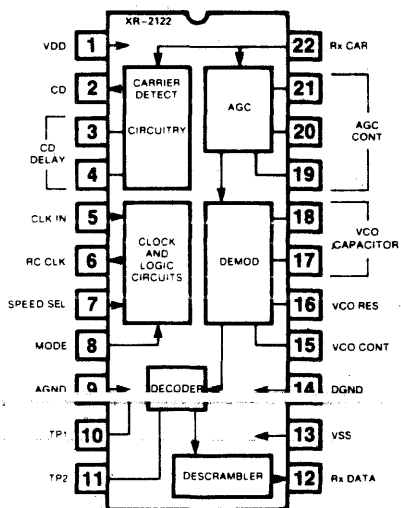
## Modem Circuits

### XR-2122 BELL 212A DEMODULATOR

The XR-2122 is designed to provide the complete demodulator function for the Bell Standard 212A PSK Modem. It can be used with the XR-2120 PSK Filter, XR-2121 Modulator, and XR-2125 Sync/Async Converter to provide a four chip solution for realizing this 1200/300 bps Modem System.

The XR-2122 Demodulator provides the complete demodulation function for either 1200 bps PSK or 0 to 300 bps FSK incoming carriers. An internal descrambler, automatic gain control circuit, and carrier detect circuit are also provided. The XR-2122 is designed to operate from dual 5-volt power supplies and utilizes CMOS technology for low power consumption.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Bell Standard 212A Compatible
- Automatic Speed Selection in Answer Mode
- 36 dB (-9 to -45 dBm) Dynamic Input Range
- On-Board Descrambler
- Carrier Detect Output
- Crystal Controlled Operation
- CMOS/TTL Compatible Inputs and Outputs

#### APPLICATIONS

- Stand-Alone Modems
- Remote Terminals
- Built-In Modems

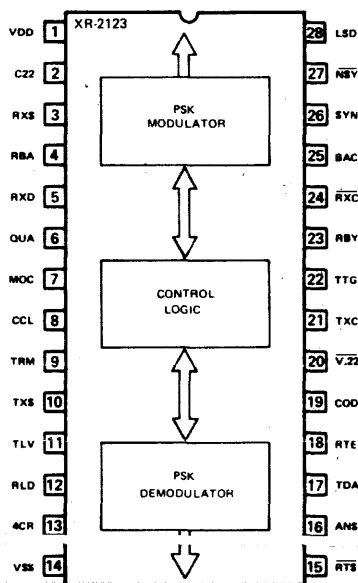
#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2122CN	Ceramic	0°C to +70°C
XR-2122CP	Plastic	0°C to +70°C

### XR-2123 PSK MODULATOR/DEMODULATOR

The XR-2123 provides the phase-shift keying (PSK) Modulator and Demodulator Functions for implementing a Full-Duplex 1200 bps or Half-Duplex 2400 bps Modem System. Using fully digital circuit techniques allows the XR-2123 to be externally programmed for operation for Bell Standards 201B and C or 212A (1200 bps only), and CCITT V.22 or V.26. Internal logic and timing functions minimize external parts while crystal controlled operation provides stable and accurate operation.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- CMOS Technology
- Single +5 Volt Operation
- Low Power Consumption (Typ 10mw)
- 1200 bps Full-Duplex
- 2400 bps Half-Duplex
- Programmable for US or European Standards (CCITT)
- Dibit PSK (DPSK) Operation
- Crystal Controlled
- Synthesized Sine Wave Modulator Output
- Adjustable Modulator Output
- Input Protection

#### APPLICATIONS

- Bell Standard 201 or 212A Modems
- CCITT Standard V.22 or V.26 Modems

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2123CN	Ceramic	0°C to +70°C
XR-2123CP	Plastic	0°C to +70°C

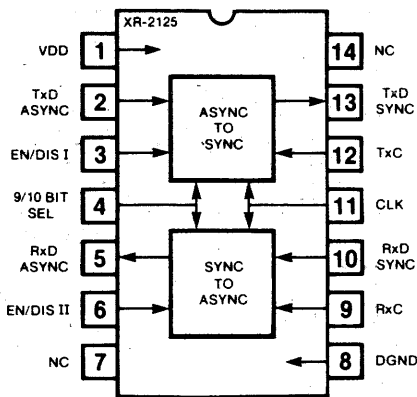
## Modem Circuits

### XR-2125 SYNC/ASYNC CONVERTER

The XR-2125 is designed to provide the interface between asynchronous digital equipment and the synchronous operation of the modulator and demodulator functions in a Bell Standard 212A PSK Modem. It can be used with the XR-2120 PSK Filter, XR-2121 Modulator, and XR-2122 Demodulator to provide a four chip solution for realizing this 1200/300 bps Modem System.

The XR-2125 maintains proper timing of data by inserting or removing stop bits when interfacing with asynchronous digital equipment. Enable/Disable pins are provided to allow straight-through operation when interfacing with synchronous equipment. Operation with character lengths of either 9 or 10 bits may be selected also. The XR-2125 is designed to operate from a single 5-volt supply and utilizes CMOS technology for low power consumption.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Bell Standard 212A Compatible
- Enable/Disable Inputs
- Crystal Controlled Operation
- CMOS/TTL Compatible Inputs
- 9 or 10 Bit Selection

### APPLICATIONS

- Stand-Alone Modems
- Remote Terminals
- Built-In Modems

### ORDERING INFORMATION

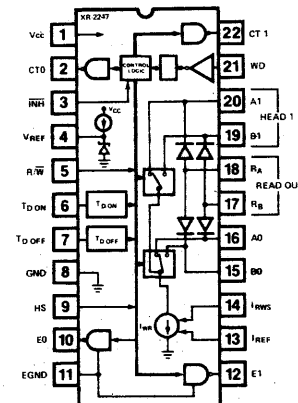
Part Number	Package	Operating Temperature
XR-2125CN	Ceramic	0°C to +70°C
XR-2125CP	Plastic	0°C to +70°C

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### XR-2247 FLOPPY DISK WRITE AMPLIFIER

The XR-2247 is a write amplifier designed to provide the complete interface between write data signals and tunnel-erase magnetic heads. Although primarily intended for floppy disk drive systems, the XR-2247 can also be used in other magnetic media systems such as tape drives. Complete internal active transistor switching is provided to minimize external parts for dual head systems. Write and erase currents are each externally programmable with a single resistor. Also included is circuitry for inner track write current compensation. To prevent false write current outputs on power-on, an inhibit input has been provided. Turn-on and turn-off times for the tunnel-erase systems are each externally programmable.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Fully Programmable Write and Erase Currents
- Fully Programmable Tunnel Delay for Turn-on/Turn-off
- Internal Switching Diodes (active transistors) for Dual Head Drives
- Single Supply Operation
- Inner Track Write Current Compensation
- Inhibit Input
- TTL Compatible Inputs
- Low External Parts Count

### APPLICATIONS

- Floppy Disk Drives
- Single/Double Head Systems
- Magnetic Tape Write Amplifier

### ORDERING INFORMATION

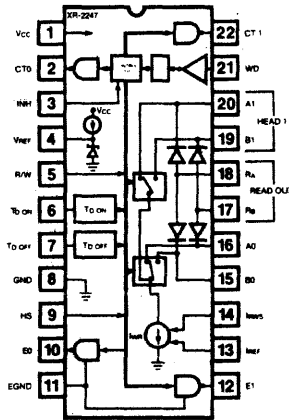
Part Number	Package	Operating Temperature
XR-2247CN	Ceramic	0°C to +70°C
XR-2247CP	Plastic	0°C to +70°C

## Disk Drive Circuits

### XR-2247A FLOPPY DISK WRITE AMPLIFIER

The XR-2247A is a write amplifier designed to provide the complete interface between write data signals and tunnel-erase magnetic heads. Although primarily intended for floppy disk drive systems, the XR-2247A can also be used in other magnetic media systems such as tape drives. Complete internal diode connected transistor switching is provided to minimize external parts for dual head systems. Write and erase currents are each externally programmable with a single resistor. Also included is circuitry for inner-track write current compensation. To prevent false write current outputs on power-on, an inhibit input has been provided. Turn-on and turn-off times for the tunnel-erase systems are each externally programmable.

#### FUNCTIONAL BLOCK DIAGRAM

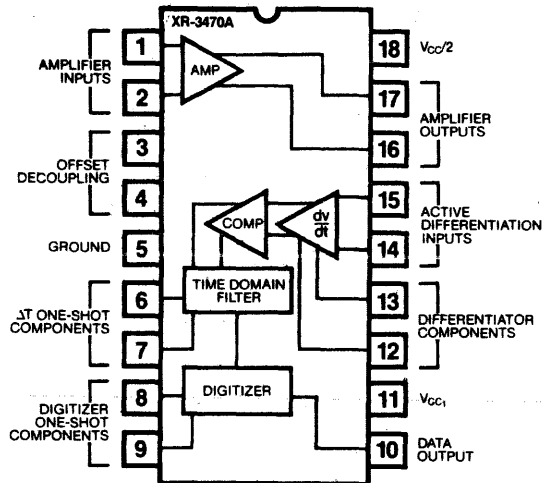


### XR-3470A FLOPPY DISK READ AMPLIFIER

The XR-3470A is a read amplifier system designed primarily for use in a floppy disk system. It is designed to perform the complete readback function, by accepting the readback signal from a magnetic head and converting it into digital output pulses. To perform this function, the XR-3470A contains a high frequency amplifier, an active differentiator, a zero-crossing detector, and a time domain filter.

The XR-3470A is suited for systems with data transfer rates up to 3 megabaud. High input sensitivity allows operation with signal levels as low as 1.4 mV pp, which gives it the flexibility to be used for single or double density floppy disk systems.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Complete Floppy Disk Read Amplifier
- Low Input Voltage Detection 1.4 mV pp
- Low Peak Shift 2.0% Maximum
- Low Amplifier Gain Variation 100 V/V Min
- High Amplifier Frequency Response 130 V/V Max
- High Amplifier Frequency Response 10 MHz, Min

#### APPLICATIONS

- Single/Double Density Floppy Disk Read Amplifier
- Magnetic Read Amplifier

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-3470ACN	Ceramic	0°C to +70°C
XR-3470ACP	Plastic	0°C to +70°C

#### FEATURES

- Fully Programmable Write and Erase Currents
- Fully Programmable Tunnel Delay for Turn-on/Turn-off
- Internal Switching Diodes (Diode connected transistors) for Dual Head Drives
- Single Supply Operation
- Inner track Write Current Compensation
- Inhibit Input
- TTL Compatible Inputs
- Low External Parts Count

#### APPLICATIONS

- Floppy Disk Drives
- Single/Double Head Systems
- Magnetic Tape Write Amplifier

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2247ACN	Ceramic	0°C to +70°C
XR-2247ACP	Plastic	0°C to +70°C

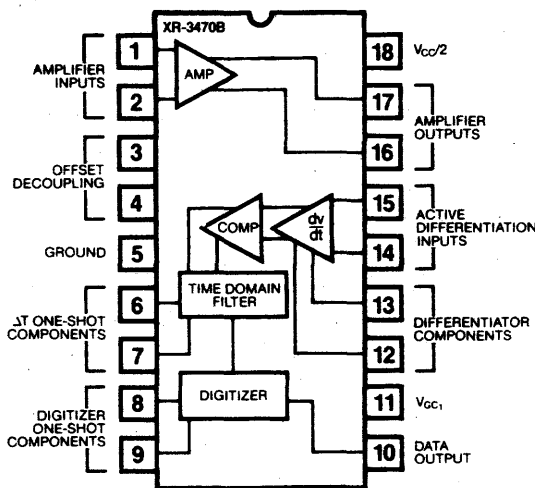
## Disk Drive Circuits

### XR-3470B FLOPPY DISK READ AMPLIFIER

The XR-3470B is a read amplifier system designed primarily for use in a floppy disk system. The circuit is designed to accept the readback signal from a magnetic head and convert the peaks of this signal to digital pulses. The circuit consists of a high-frequency amplifier, an active differentiator, a zero-crossing detector, and a time domain filter.

The XR-3470B can be used to transfer data up to 3 megabaud. This device can also detect voltage levels as low as 1.4 mV pp, which gives it the flexibility to be used for single or double density floppy disk systems. The chip contains all the circuitry essential for a floppy disk read amplifier system.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Less Than 4% Peak Shift
- Low Input Voltage Detection
- Complete Floppy Disk Read System
- Up to 3 MBaud Operation
- High Preamp Gain

110 V/V, Typical,  
(100 V/V, min)

#### APPLICATIONS

Single/Double Density Floppy Disk Read Amplifier  
Magnetic Tape Read Amplifier

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-3470BCN	Ceramic	0°C to +70°C
XR-3470BCP	Plastic	0°C to +70°C

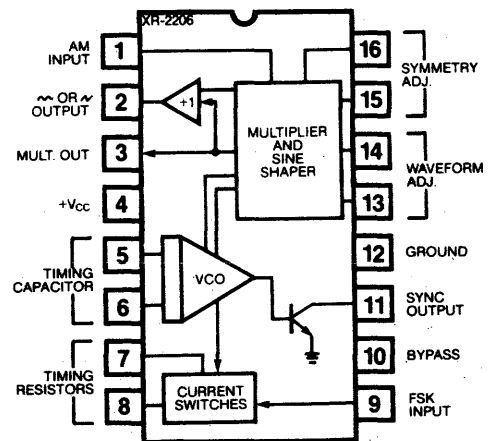
## Function Generators

### XR-2206 MONOLITHIC FUNCTION GENERATOR

The XR-2206 is a monolithic function generator integrated circuit capable of producing high quality sine, square, triangle, ramp, and pulse waveforms of high-stability and accuracy. The output waveforms can be both amplitude and frequency modulated by an external voltage. Frequency of operation can be selected externally over a range of 0.01 Hz to more than 1 MHz.

The circuit is ideally suited for communications, instrumentation, and function generator applications requiring sinusoidal tone, AM, FM, or FSK generation.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Low-Sine Wave Distortion: .5%, Typical
- Excellent Stability: 20 ppm/°C, Typical
- Wide Sweep Range: 2000:1, Typical
- Linear Amplitude Modulation
- Adjustable Duty Cycle: 1% to 99%

#### APPLICATIONS

- Waveform Generation: V/F Conversion
- Sweep Generation: FSK Generation
- AM/FM Generation: Phase-Locked Loops

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2206M	Ceramic	-55°C to +125°C
XR-2206N	Ceramic	0°C to +70°C
XR-2206P	Plastic	0°C to +70°C
XR-2206CN	Ceramic	0°C to +70°C
XR-2206CP	Plastic	0°C to +70°C

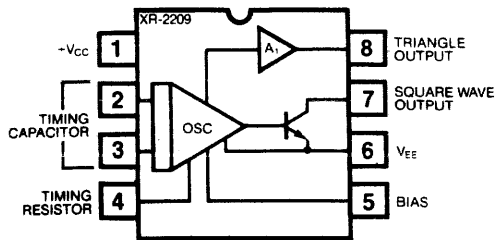
## Function Generators

### XR-2209 PRECISION OSCILLATOR

The XR-2209 is a monolithic variable frequency oscillator circuit featuring excellent temperature stability and a wide linear sweep range. The circuit provides simultaneous triangle and square wave outputs over a frequency range of 0.01 Hz to 1 MHz. The frequency is set by an external RC product. The device is ideally suited for frequency modulation, voltage-to-frequency or current-to-frequency conversion, sweep or tone generation, as well as for phase-locked loop applications when used in conjunction with a phase comparator such as the XR-2208.

The circuit is comprised of three functional blocks: A variable frequency oscillator and two buffer amplifiers. The XR-2209 has a typical drift specification of 20 ppm/°C. The frequency can be linearly swept over a 1000:1 range, by an external control signal.

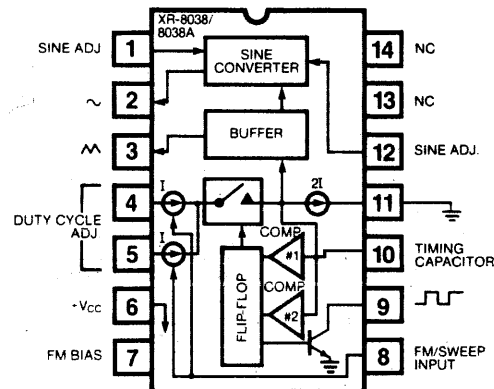
#### FUNCTIONAL BLOCK DIAGRAM



### XR-8038/8038A PRECISION WAVEFORM GENERATOR

The XR-8038/8038A is a precision waveform generator IC capable of producing sine, square, triangular, sawtooth, and pulse waveforms with a minimum of external components and adjustments. Its operating frequency can be selected over nine decades of frequency, from 0.001 Hz to 1 MHz, by the choice of external RC components. The frequency of oscillation is highly stable over a wide temperature and supply voltage range. The frequency modulation and sweeping can be accomplished with an external control voltage, without affecting the quality of the output waveforms. The frequency can be programmed digitally through the use of either resistors or capacitors.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Direct Replacement for Intersil 8038	
Low Frequency Drift	50 ppm/°C
Simultaneous Sine, Triangle and Square Wave Outputs	
Low Distortion	THD 1%
High FM and Triangle Linearity	
Wide Frequency Range	0.001 Hz to 1 MHz
Minimum External Component Count	
Variable Duty Cycle	2% to 98%

#### APPLICATIONS

Precision Waveform Generation	Sweep Generation
Test Instrumentation Design	Precision PLL Design
Phase-Locked Clock Generation	Tone Generation

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-8038M/AM	Ceramic	-55°C to +125°C
XR-8038N/AN	Ceramic	0°C to +70°C
XR-8038P/AP	Plastic	0°C to +70°C
XR-8038CN/ACN	Ceramic	0°C to +70°C
XR-8038CP/ACP	Plastic	0°C to +70°C

#### FEATURES

Excellent Temperature Stability	20 ppm/°C
Linear Frequency Sweep	
Wide Sweep Range	1000:1, Minimum
Wide Supply Voltage Range	±4V to ±13V
Low Supply Sensitivity	0.15%/V
Wide Frequency Range	0.01 Hz to 1 MHz
Simultaneous Triangle and Square Wave Outputs	

#### APPLICATIONS

V/F Conversion	Waveform Generation
C/F Conversion	Stable Phase-Locked Loop
FM and Sweep Generation	

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2209M	Ceramic	-55°C to +125°C
XR-2209CN	Ceramic	0°C to +70°C
XR-2209CP	Plastic	0°C to +70°C

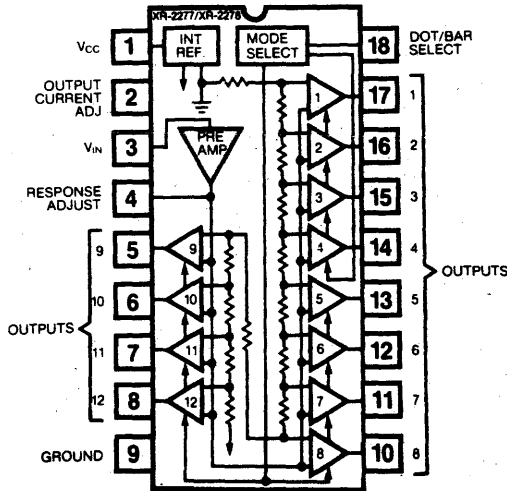
## Display Drivers

### XR-2277/XR-2278 DOT AND BAR-GRAPH DISPLAY GENERATORS

The XR-2277 and XR-2278 are 12-point level detector circuits designed for interfacing directly with LED moving dot or bar-graph displays. Each circuit is comprised of an input buffer amplifier, a set of 12 comparators which are biased from an internal voltage reference, and a resistor string. Each comparator provides a high-impedance current source output; each of the output currents is very closely matched and can be adjusted, simultaneously, with a single external setting resistor.

The XR-2277 has 12 discrete output levels, over a range of -30 dB to +6 dB, referenced to an externally set zero dB reference level. The XR-2278 has similar electrical characteristics, except for a dynamic range of -20 dB to +8 dB.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Direct LED Interface
- Adjustable Zero dB Reference
- Constant-Current Outputs
- Adjustable Output Currents
- External Mode-Select
- High Input Impedance
- for Dot/Bar-Graph Format

#### APPLICATIONS

- Bar-Graph Display Generator and Driver
- Dot Display Generator

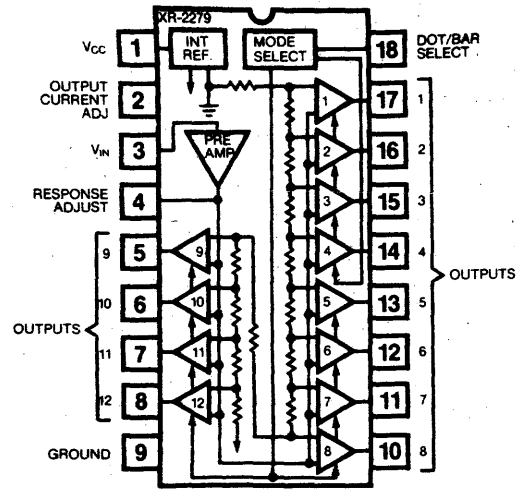
#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2277CP	Plastic	0°C to +70°C
XR-2278CP	Plastic	0°C to +70°C

### XR-2279 DOT AND BAR-GRAPH DISPLAY GENERATOR

The XR-2279 is a 12-point logarithmic level-detector circuit comprised of an input buffer amplifier and a set of 12 voltage comparators. The circuit produces 12 discrete output levels, spaced in three dB intervals, over a dynamic range of -27 dB to +6 dB, referenced to an externally adjusted zero dB level. It is designed for interfacing directly with LED moving dot or bar-graph displays. Each of the comparator outputs provides a high-impedance constant-current drive which is well-matched and can be adjusted by an external resistor setting. The circuit output is either a moving dot or a continuous bar-graph format, based on control voltage applied to the mode-select pin.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Direct LED Interface
- Constant 3 dB
- per Step Logarithmic Scale
- External Mode-Select
- for Dot/Bar-Graph Formats
- Adjustable Output Current Levels
- Adjustable Zero dB Reference

#### APPLICATIONS

- Bar-Graph Generator
- Logarithmic Level Indicator
- Moving Dot Display Generator
- Sequential Level Indicator

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-2279CP	Plastic	0°C to +70°C

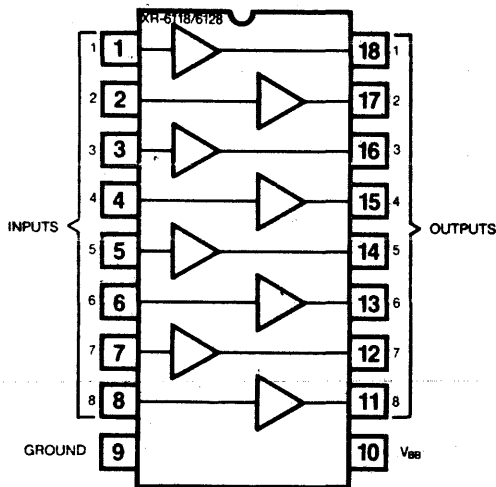
## Display Drivers

## Operational Amplifiers

### XR-6118/6128 FLUORESCENT DISPLAY DRIVER

The XR-6118 and XR-6128 are high-voltage display driver arrays which are designed to interface between low-level digital logic and vacuum fluorescent displays. Each circuit consists of eight independent signal channels comprised of Darlington output stages and common-emitter inputs. All channels on the chip share common power supply and ground connections. Both device types are capable of driving digits and/or segments of fluorescent displays, and all of the eight outputs can be activated simultaneously. The XR-6118 is compatible with TTL and 5V CMOS logic while the XR-6128 interfaces with 6V to 15V PMOS or CMOS. The XR-6118 and XR-6128 are designed to operate with a supply voltage of 80V, and the XR-6118-2 with a supply voltage of 60V.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Direct Replacement for Sprague 6118A/6128A  
 Digit or Segment Drivers  
 Low Power and Input Current  
 Internal Output Pulldown Resistors  
 High Output Breakdown Voltage 77V, Minimum

#### APPLICATIONS

Fluorescent Driver High-Voltage Switching  
 Gas Discharge Display Driver

#### ORDERING INFORMATION

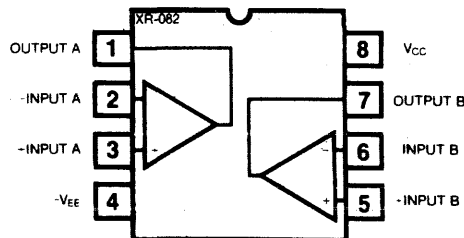
Part Number	Package	Operating Temperature
XR-6118P	Plastic	-40°C to +85°C
XR-6128CP	Plastic	-40°C to +85°C
XR-6118CP	Plastic	0°C to +70°C
XR6128CP	Plastic	0°C to +70°C
XR-6128CP	Plastic	0°C to +70°C
XR-6118CP-2	Plastic	0°C to +70°C

### XR-082/083 DUAL BIPOLAR JFET OPERATIONAL AMPLIFIER

The XR-082/083 family of junction FET input dual operational amplifiers are designed to offer higher performance than conventional bipolar op amps. Each amplifier features high slew rate, low input bias and offset currents, and low offset voltage drift with temperature. These operational amplifier circuits are fabricated using ion implantation technology to combine well-matched junction FET's and high-performance bipolar transistors on the same monolithic chip.

The XR-082 family of dual bipolar JFET op amps are packaged in an 8-Pin dual-in-line package. The XR-083 family of op amps offers an independent offset adjustment for each of the individual op amps on the same chip, and are available in a 14-Pin dual-in-line package.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Direct Replacement for TL082/083 (See Chart)  
 Low Power Consumption  
 Wide Common-Mode and Differential Voltage Ranges  
 Low Input Bias and Offset Currents  
 Output Short Circuit Protection  
 High-Input Impedance . . . . FET Input Stage  
 Internal Frequency Compensation  
 Latch-Up-Free Operation  
 High Slew Rate 13V/ $\mu$ s, Typical

#### APPLICATIONS

Transducer Amplifier High-Impedance Buffer

#### ORDERING INFORMATION

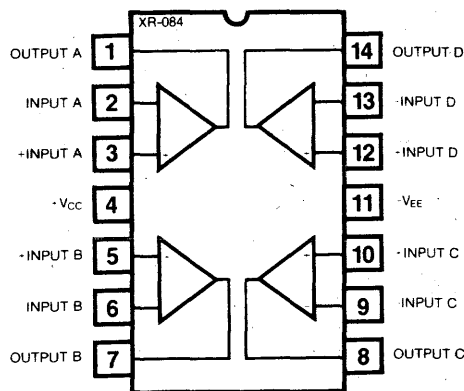
Part Number	Package	Operating Temperature
XR-082M	Ceramic	-55°C to +125°C
XR-082N	Ceramic	-25°C to +85°C
XR-082P	Plastic	-25°C to +85°C
XR-082CN	Ceramic	0°C to +70°C
XR-082CP	Plastic	0°C to +70°C
XR-083	Available in 14-Pin	

## Operational Amplifiers

### XR-084 QUAD BIPOLAR JFET OPERATIONAL AMPLIFIER

The XR-084 junction FET input quad operational amplifier is designed to offer higher performance than conventional bipolar quad op amps. Each of the four op amps on the chip is closely matched in performance characteristics, and each amplifier features high slew rate, low input bias and offset currents, and low offset voltage drift with temperature. The XR-084 FET input quad op amp is fabricated using ion implantation bipolar/FET technology, to combine well-matched junction FET's and high-performance bipolar transistors on the same monolithic chip.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Direct Replacement for TL084 (See Table 1)
- Same Pin Configuration as XR-3403 and LM324
- High-Impedance Junction FET Input Stage
- Low-Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Internal Frequency Compensation
- Latch-Up Free Operation
- High Slew Rate 13V/ $\mu$ s, Typical

#### APPLICATIONS

Transducer Amplifier High-Impedance Buffer

#### ORDERING INFORMATION

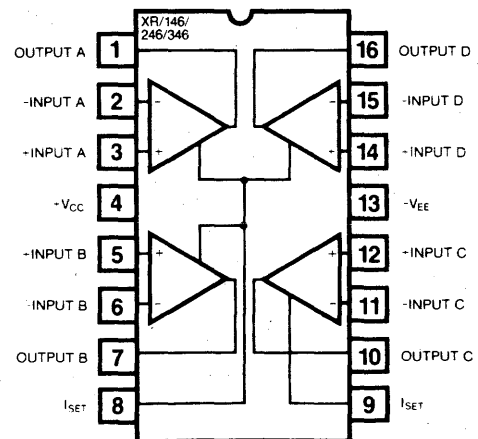
Part Number	Package	Operating Temperature
XR-084M	Ceramic	-55°C to +125°C
XR-084N	Ceramic	-25°C to +85°C
XR-084P	Plastic	-25°C to +85°C
XR-084CN	Ceramic	0°C to +70°C
XR-084CP	Plastic	0°C to +70°C

### XR-146/246/346 PROGRAMMABLE QUAD OPERATIONAL AMPLIFIER

The XR-146 family of quad operational amplifiers contain four independent high-gain, low-power, programmable op amps on a monolithic chip. The use of external bias setting resistors permit the user to program the gain bandwidth product, supply current, input bias current, input offset current, input noise, and slew rate.

The basic XR-146 family of circuits offers partitioned programming of the internal op amps. One setting resistor is used to set the bias levels in the three op amps, and a second bias setting is used for the remaining op amp. Its modified version, XR-346-2, provides a separate bias setting resistor for each of the two op amp pairs.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Direct Replacement for LM146/246/346
- Direct Replacement for LM346-2
- Programmable Electrical Characteristics
- Low Supply Current 350  $\mu$ A per Amplifier
- Large dc Voltage Gain 120 dB
- Low Noise Voltage 25 nV/ $\sqrt$ Hz
- Wide Power Supply Range  $\pm$  1.5V to  $\pm$  22V
- Class AB Output Stage No Crossover Distortion

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-146M	Ceramic	-55°C to +125°C
XR-246N	Ceramic	-25°C to +85°C
XR-246P	Plastic	-25°C to +75°C
XR-346CN	Ceramic	0°C to +70°C
XR-346CP	Plastic	0°C to +70°C
XR-346-2CN	Ceramic	0°C to +70°C
XR-346-2CP	Plastic	0°C to +70°C

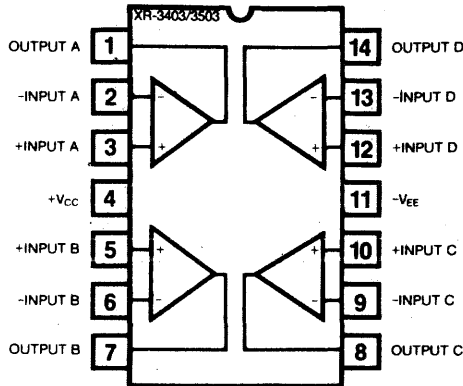


# Operational Amplifiers

## XR-3403/3503 QUAD OPERATIONAL AMPLIFIER

The XR-3403 is an array of four independent operational amplifiers, each with true differential inputs. The device has electrical characteristics similar to the popular 741. However, the XR-3403 has several distinct advantages over standard operational amplifier types in single supply applications. The XR-3403 can operate at supply voltages as low as 3.0 volts, or as high as 36 volts. Quiescent currents remain about one-fifth of those associated with the 741 (on a per amplifier basis). The common-mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications. The output voltage swing extends to the negative supply.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Direct Pin for Pin Replacement for MC3403/3503, LM324 & RC4137
- Suitable for Single Supply Operation
- Short Circuit Protected Outputs
- Class AB Output Stage
- Single Supply Operation
- Split Supply Operation
- Low Input Bias Currents
- Internally Compensated
- No Crossover Distortion
- 3.0V to 36V
- $\pm 1.5V$  to  $\pm 18V$
- 500 nA Maximum

### APPLICATIONS

- Circuits Requiring Ground Sensing at Inputs
- Low Supply Voltage Systems

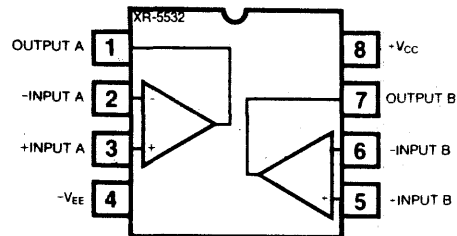
### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-3503M	Ceramic	-55°C to +125°C
XR-3403CN	Ceramic	0°C to +70°C
XR-3403CP	Plastic	0°C to +70°C

## XR-5532/5532A DUAL LOW-NOISE OPERATIONAL AMPLIFIER

The XR-5532 ultralow noise dual operational amplifier is especially designed for high quality audio and instrumentation applications. Compared to standard 741- or 301A-type op amps, these operational amplifiers offer an order of magnitude improvement in both noise performance and small signal bandwidth. Higher output drive capability and internal compensation are added features. The XR-5532A is specially screened for a guaranteed ultralow noise specification. This ultralow noise specification makes the XR-5532A ideally suited for audio preamplifier and other low level applications.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Pin for Pin Replacement MC1458, RC4558, TL072 TL082 & LF 353
- Direct Replacement for NE5532/NE5532A
- Ultralow Input Noise
- Wide Small Signal Bandwidth
- High Output Drive Capability
- Wide Supply Range
- Wide Power Bandwidth
- High Slew Rate
- 4 nV/√Hz, Typical
- 10 MHz, Typical
- 10V rms into 600Ω
- $\pm 3V$  to  $\pm 20V$
- 200 kHz
- 6V/μsec

### APPLICATIONS

- Professional Audio Equipment
- Instrumentation and Servo Control
- Telephone Channel Amplifier
- Low-Level Signal Processing
- Audio Preamplification
- Transducer Amplifiers

### ORDERING INFORMATION

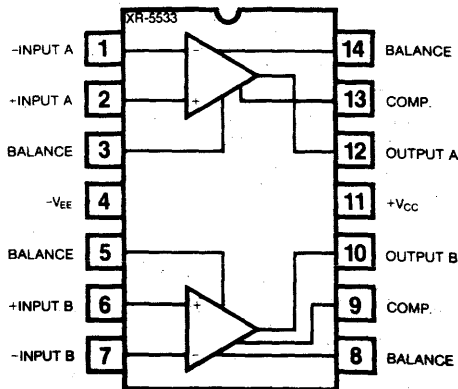
Part Number	Package	Operating Temperature
XR-5532ACN	Ceramic	0°C to +70°C
XR-5532CN	Ceramic	0°C to +70°C
XR-5532ACP	Plastic	0°C to +70°C
XR-5532CP	Plastic	0°C to +70°C

## Operational Amplifiers

### XR-5533/5533A DUAL LOW-NOISE OPERATIONAL AMPLIFIER

The XR-5533 ultralow noise dual operational amplifier is especially designed for high quality audio and instrumentation applications. Compared to the standard 741- or 301A-type op amps, these operational amplifiers offer an order of magnitude improvement in noise performance, and small signal bandwidth, in addition to higher output drive capability. The XR-5533 is internally compensated for a voltage gain of three or more, and offers independent offset adjustments for each of the two op amps. The XR-5533A is specially screened for a guaranteed ultralow noise specification.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Direct Replacement for NE5533/NE5533A  
 Dual Version of XR-5534 or NE5534  
 Ultralow Input Noise 4 nV/√Hz, Typical  
 Wide Small Signal Bandwidth 10 MHz, Typical  
 High Output Drive Capability 10V rms into 600Ω  
 Wide Supply Range ±3V to ±20V  
 Wide Power Bandwidth 200 kHz  
 High Slew Rate 13V/μsec

#### APPLICATIONS

Professional Audio Equipment  
 Instrumentation and Servo Control  
 Telephone Channel Amplifier  
 Low-Level Signal Processing  
 Audio Pre amplification

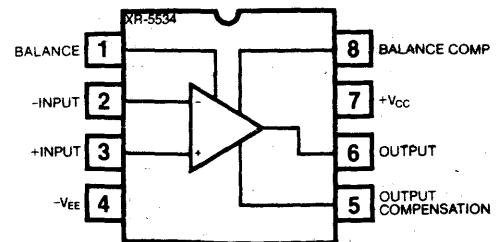
#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-5533ACN	Ceramic	0°C to +70°C
XR-5533CN	Ceramic	0°C to +70°C
XR-5533ACP	Plastic	0°C to +70°C
XR-5533CP	Plastic	0°C to +70°C

### XR-5534/5534A LOW-NOISE OPERATIONAL AMPLIFIER

The XR-5534 is a high performance, ultralow noise operational amplifier. Compared to the standard 741- or 301A-type op amps, it offers an order of magnitude improvement in noise performance, and small signal bandwidth, as well as significantly better output drive capability. The XR-5534 is ideally suited for applications in high quality and professional audio equipment, instrumentation, control circuits, and telephone channel amplifiers. The op amp is internally compensated for a gain of three or greater. The frequency response can be optimized with an external compensation capacitor for applications requiring unity-gain, low-overshoot response, or capacitive load driving. The XR-5534A is specially screened for a guaranteed ultralow noise specification.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Direct Replacement for NE5534/NE5534A  
 Ultralow Input Noise 4 nV/√Hz, Typical  
 Wide Small Signal Bandwidth 10 MHz, Typical  
 High Output Drive Capability 10V rms into 600Ω  
 Wide Supply Range ±3V to ±20V  
 High-Voltage Gain Av = 100,000 at dc  
 High Slew Rate 13V/μsec

#### APPLICATIONS

Professional Audio Equipment  
 Audio Pre amplification

#### ORDERING INFORMATION

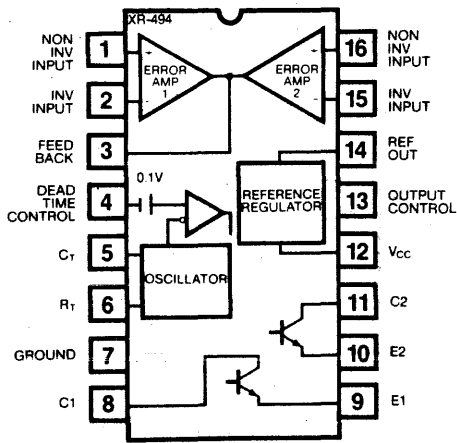
Part Number	Package	Operating Temperature
XR-5534AM	Ceramic	-55°C to +125°C
XR-5534M	Ceramic	-55°C to +125°C
XR-5534ACN	Ceramic	0°C to +70°C
XR-5534CN	Ceramic	0°C to +70°C
XR-5534ACP	Plastic	0°C to +70°C
XR-5534CP	Plastic	0°C to +70°C

# Voltage Regulators

## XR-494 PULSE-WIDTH MODULATING REGULATOR

All functions required to construct a pulse-width modulating regulator are incorporated on a single monolithic chip in the XR-494. The device is primarily designed for power supply control and contains an on-chip 5-volt regulator, two error amplifiers, an adjustable oscillator, a dead-time control comparator, a pulse-steering flip-flop, and output control circuits. Either common emitter or emitter follower output capability is provided by the uncommitted output transistors.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Complete PWM Power Control Circuitry
- Uncommitted Outputs for 200 mA Sink or Source
- Output Control Selects Single-ended or Push-pull Operation
- Internal Circuitry Prohibits Double Pulse at Either Output
- Variable Dead-time Provides Control Over Total Range
- Internal Regulator Provides a Stable 5V Reference Supply
- Circuit Architecture Provides Easy Synchronization

### APPLICATIONS

- Pulse-Width Modulated Power Control Systems
- Switching Regulators
- Motor Speed Control
- Power Inverters

### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-494M	Ceramic	-55°C to +125°C
XR-494CN	Ceramic	0°C to +70°C
XR-494CP	Plastic	0°C to +70°C

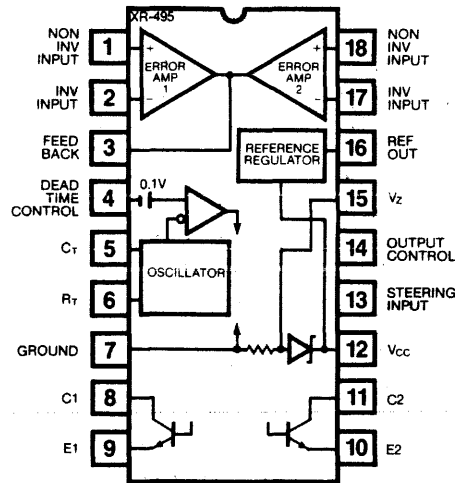
3380

## XR-495 PULSE-WIDTH MODULATING REGULATOR

All functions required to construct a pulse-width modulating regulator are incorporated on a single monolithic chip. The device is primarily designed for power supply control and contains an on-chip 5-volt regulator, two error amplifiers, an adjustable oscillator, a dead-time control comparator, a pulse-steering flip-flop, and output control circuits. Either common emitter or emitter follower output capability is provided by the uncommitted output transistors.

The XR-495 also contains an on-chip 39-volt Zener diode for high-voltage applications, where  $V_{CC}$  is greater than 40 volts, and an output steering control that overrides the internal control of the pulse-steering flip-flop.

### FUNCTIONAL BLOCK DIAGRAM



### FEATURES

- Complete PWM Power Control Circuitry
- Uncommitted Outputs for 200 mA Sink or Source
- Output Control Selects Single-ended or Push-pull Operation
- Internal Circuitry Prohibits Double-Pulse at Either Output
- Variable Dead-time Provides Control Over Total Range

### APPLICATIONS

- Pulse-Width Modulated Power Control Systems
- Switching Regulators

### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-495M	Ceramic	-55°C to +125°C
XR-495CN	Ceramic	0°C to +70°C
XR-495CP	Plastic	0°C to +70°C

LINEAR  
Exar Integrated Systems

## Line Interface Circuits

### XR-1488 QUAD LINE DRIVER

The XR-1488 is a monolithic quad line driver designed to interface data terminal equipment with data communications equipment, in conformance with the specifications of EIA Standard No. RS232C. This extremely versatile integrated circuit can be used to perform a wide range of applications. The circuit features output current limiting circuitry, and independent positive and negative power supply driving elements. Compatibility with all DTL and TTL logic families enhances the versatility of the circuit.

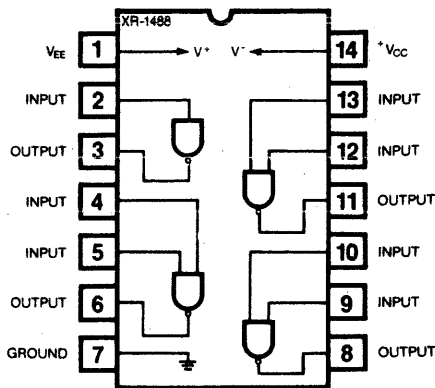
The XR-1488 quad line driver along with its companion circuit, the XR-1489A quad line receiver, provides a complete interface system between DTL and TTL logic levels, and the RS232C defined voltage and impedance levels.

### XR-1489A QUAD LINE RECEIVER

The XR-1489A is a monolithic quad line receiver especially designed for data bus interface. Each of the line receiver sections has adjustable hysteresis characteristics for improved noise rejection. The input and output levels of the circuit are designed to provide direct interface between RS232C data bus standards and the DTL or TTL type logic levels.

The XR-1489A quad line receiver along with its companion circuit, the XR-1488 quad line driver, provides a complete interface between DTL and TTL logic levels, and the RS232C defined voltage and impedance levels.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Current Limited Output
- Independent Positive Power Supply Driving Elements
- Independent Negative Power Supply Driving Elements
- Compatible with DTL and TTL Logic Families
- Data Terminal/Data Communication Interface
- Conforms to EIA Standard No. RS232C

#### APPLICATIONS

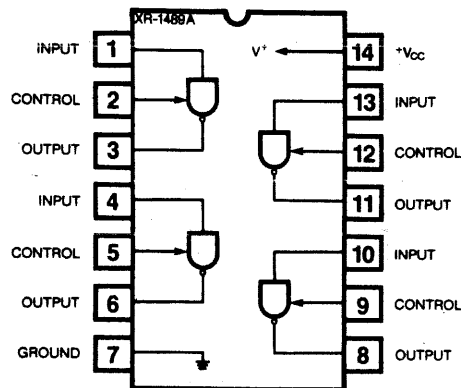
RS232 C Interface

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-1488N	Ceramic	0°C to +70°C
XR-1488P	Plastic	0°C to +70°C

3384

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Direct Replacement for MC1489A
- Current Limited Output
- Compatible with DTL and TTL Logic
- Meets EIA Standard RS232C

#### APPLICATIONS

- Data Bus Interface
- Microprocessor Interface
- Remote Terminal Interface
- RS232 Interface

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-1489AN	Ceramic	0°C to +70°C
XR-1489AP	Plastic	0°C to +70°C

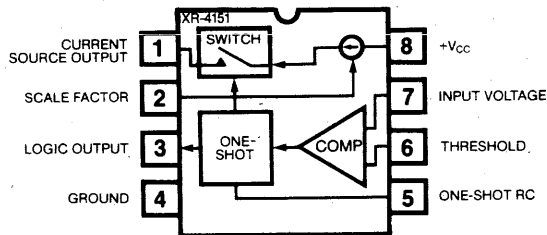
© IC MASTER 1984

## Special Functions

### XR-4151 VOLTAGE-TO-FREQUENCY CONVERTER

The XR-4151 is a device designed to provide a simple, low cost, method for converting a dc voltage into a proportional pulse repetition frequency. It is also capable of converting an input frequency into a proportional output voltage. The XR-4151 is useful in a wide range of applications, including A/D and D/A conversion and data transmission.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

Single Supply Operation	+ 8V to + 22V
Pulse Output Compatible with all Logic Forms	
Programmable Scale Factor	
Linearity (Precision Mode)	± 0.05%, Typical
Temperature Stability	± 100 ppm/°C, Typical
High-Noise Rejection	
Inherent Monotonicity	
Easily Transmittable Output	
Simple Full Scale Trim	
Single-ended Input	
Frequency-to-Voltage Conversion	
Direct Replacement for RC/RV/RM4151	

#### APPLICATIONS

- Voltage-to-Frequency Conversion
- A/D and D/A Conversion
- Data Transmission
- Frequency-to-Voltage Conversion
- Transducer Interface
- System Isolation

#### ORDERING INFORMATION

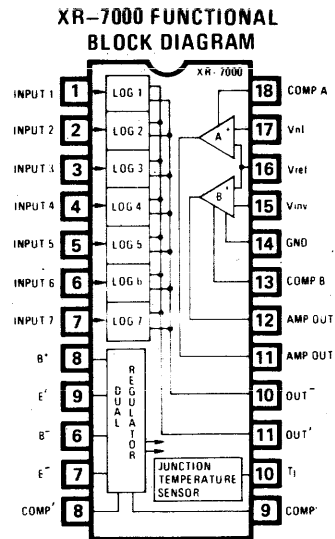
Part Number	Package	Operating Temperature
XR-4151P	Plastic	-40°C to +85°C
XR-4151CP	Plastic	0°C to +70°C

### XR-7000 LOGARITHMIC AMPLIFIER

The XR-7000 provides an internal band-gap voltage reference, a differential video summing amplifier, and a precision die temperature sensor, to aid in its system interfacing. Also included is an internal power supply to provide excellent power supply rejection.

The XR-7000, available in a 24-Pin ceramic or plastic package, is designed to operate from dual 8-volt power supplies.

#### FUNCTIONAL BLOCK DIAGRAM



#### FEATURES

- Seven Uncommitted Logging Elements
- Internal Band-Gap Voltage Reference
- Dual Tracking Regulator
- On-Board Precision Die Temperature Sensor

#### APPLICATIONS

- Receiver Subsystems
- Radar Subsystems
- Spectrum Analyzers
- Power Meters
- Test Equipment
- Video Cartridge Tape Recorders
- Audio Tape Recorders
- Smoke Detectors

#### ORDERING INFORMATION

Part Number	Package	Operating Temperature
XR-7000M	Ceramic	-55°C to +125°C
XR-7000CP	Plastic	0°C to +70°C

## Semi-Custom Development

### SEMI-CUSTOM BIPOLAR PROGRAM

The Exar bipolar semi-custom design program offers a variety of semi-custom chips to fulfill various application performance requirements and complexities.

These semi-custom chips offer a unique method of manufacturing an almost unlimited variety of custom linear and digital integrated circuits, with greatly reduced cost and development time. Exar makes this possible by stocking wafers that are completely fabricated except for the final process step of device interconnection, which metalizes the selected components together in the required circuit configurations.

### BIPOLAR SEMI-CUSTOM CHIPS

Chip Type	Chip Size	Breakdown			
	in Mils	Voltage	NPN	PNP	JFET's
A100	73 x 83	20V	60	18	
B100	85 x 85	20V	69	12	
C100A	56 x 62	20V	23	8	
D100	80 x 80	36V	50	16*	
E100	80 x 81	20V	48	15	
F100	91 x 110	20V	97	32*	
G100	90 x 90	20V	60	18*	
H100	82 x 92	20V	72	22*	
J100	61 x 65	20V	38	12*	
L100	86 x 105	20V	80	24	
M100	176 x 121	20V	149	52	
U100	**	36V	**	**	**
V100	130 x 100	36V	124	44	8
W100	130 x 160	36V	200	68	8
X100	115 x 95	75V	34	16	

\* Dual collector PNP transistors

\*\* Contact Custom Product Application Engineering for more details.

Exar offers a design kit which contains simple instructions and guidelines for designing the metal mask as well as the actual breadboard components (consisting of npn and pnp arrays and integrated resistors), which are representative of the devices available on the semi-custom chips. This provides the design engineer with the ability to closely evaluate his design's performance prior to integrating it on a monolithic chip.

### SEMI-CUSTOM I<sup>2</sup>L PROGRAM

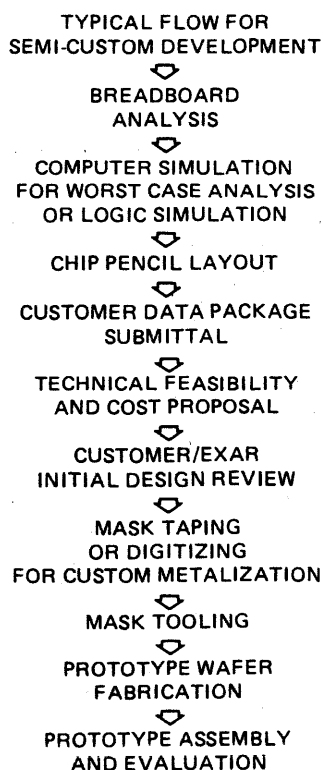
With the introduction of the I<sup>2</sup>L Gate Array chips, Exar has extended its semi-custom design program to the Integrated Injection Logic (I<sup>2</sup>L) technology. This unique method of custom LSI development technique now makes it possible to manufacture an almost unlimited variety of digital or analog/digital circuits using I<sup>2</sup>L technology, with greatly reduced development cost and time.

### I<sup>2</sup>L Semi-Custom Chips

Characteristics	CHIP SIZE			
	XR-200	XR-300	XR-400	XR-500
Chip Size (mils)	98 x 119	104 x 146	119 x 156	122 x 185
I <sup>2</sup> L Gates	192	288	256	520
Max. Operating Voltage	7V	7V	7V	7V
Bipolar I/O Interfaces	24	28	18	40
Bonding Pads	30	34	40	42

The XR-300 and XR-500 gate arrays are intended primarily for digital LSI designs. The XR-400 gate array features the advantage of combining analog and digital functions on the same IC chip. These I<sup>2</sup>L gate array chips are customized by three custom mask patterns, which are simultaneously generated from a pencil layout, using Exar's unique computerized mask generation technique. In this manner, the chip layout is greatly simplified and gate utilization efficiency is increased.

Exar also offers an I<sup>2</sup>L design kit which is intended to familiarize the designer with the basic features of I<sup>2</sup>L technology, provides helpful design guidelines in reducing his design concept to breadboard and, finally, the IC layout stage.



## Semi-Custom Development

### SEMI-CUSTOM CMOS PROGRAMS

#### Metal Gate CMOS

Exar now offers two families of CMOS Gate Arrays. The first family consists of four chips featuring from 112 to 544 uncommitted logic gates, fabricated with 7.5 $\mu$  metal gate technology.

#### CMOS Master-Chips

Chip Type	CMOS Gates	I/O Cells	Bonding Pads
XR-CMA	112	29	32
XR-CMB	162	34	38
XR-CMC	216	40	44
XR-CMD	544	46	53

A CMOS design kit is also available to familiarize the designer with the basic features of CMOS technology and to assist in the breadboarding layout of the IC chip.

#### Silicon Gate CMOS

The newest family of semi-custom Master-Chips to be added to Exar's product line is a series of four silicon gate arrays (4 $\mu$  channel length drawn approximately 3 $\mu$  actual) and range in size from 500 gates to 1600 gates. These devices feature dual layer metal and an extensive library of standard cells. Please contact Exar Custom Product Applications Engineering for more information.

### ECONOMICS OF SEMI-CUSTOM

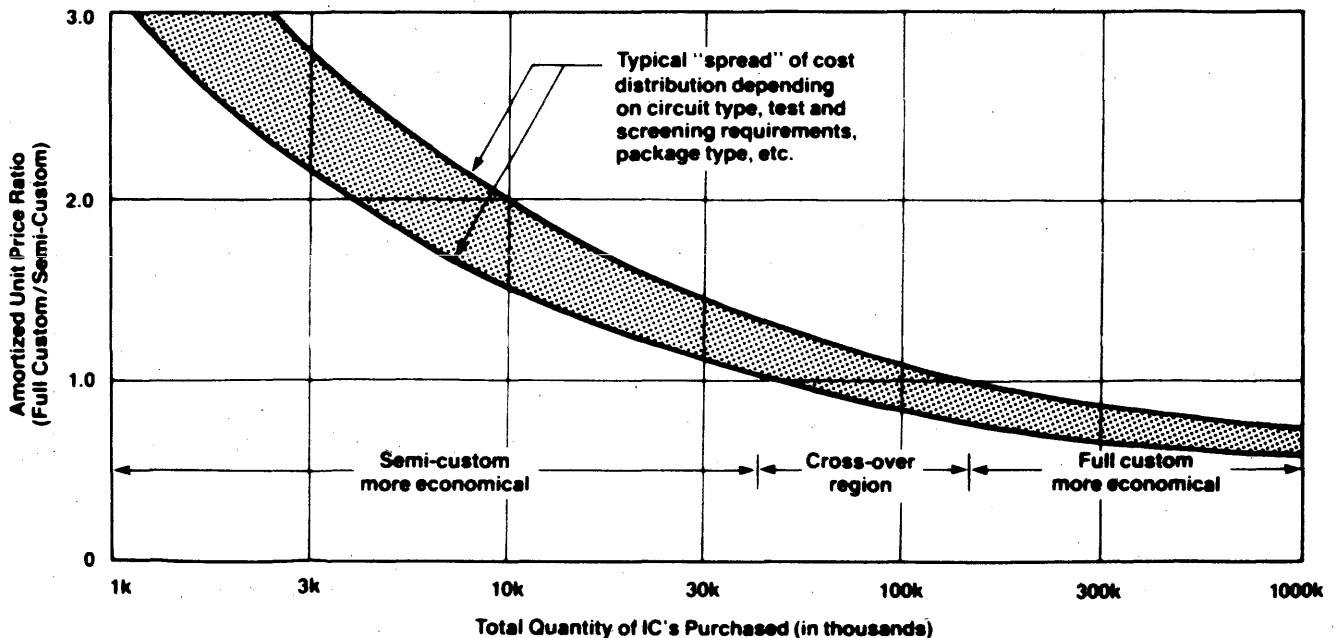
As the Amortized Unit Price Ratio graph in the figure below indicates, for production runs of 50,000 units or less, development costs are a significant influence, and a semi-custom program is clearly the best option. Over 150,000 pieces, production costs become the most important factor, and you are better off with a full custom chip.

#### Converting Semi-Custom To Full Custom

When production levels get closer to 150,000 units, converting from semi-custom to full custom is most cost effective. The conversion is simple and you get the lower production cost advantages of full custom without giving up the development benefits of semi-custom. In addition, the semi-custom prototypes often serve as a monolithic breadboard to optimize and debug the final design. The actual conversion is a relayout, rather than a redesign. The unused undedicated components are eliminated, and the circuit is laid out on a smaller chip.

#### Cost Savings Of Two-Step Conversion

Semi-custom design requires only 10-20% of the full custom development costs. Conversion from semi-custom to full custom costs about 50% of the total cost of full custom development. Therefore, the total development costs of the two-step conversion are only 60-70% of a complete full custom program.



# Processing sparks improvement of semicustom linear chips

A number of modern semiconductor processing techniques previously not used for semicustom linear ICs were implemented in the XR-W100 and XR-V100 Master Chips.

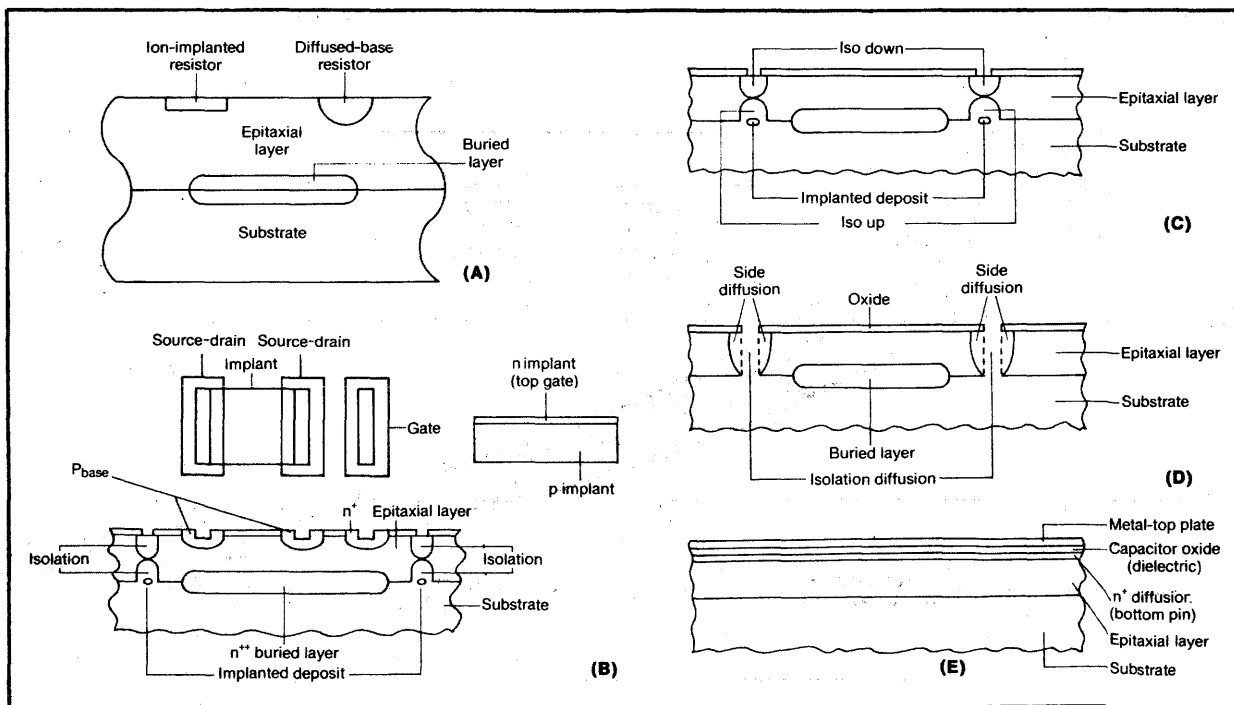
**Ion implantation.** Ion implantation ballistically introduces dopant ions into the surface of a semiconductor material. Because a machine similar to a linear accelerator shoots the dopant into the surface, their penetration and concentration can be precisely controlled by varying the particle-beam dosage and energy. This technology produces having very high sheet resistivity; thus high-value resistors can fit into much smaller areas on the chip than previously. In addition, ion-implanted resistors have about 300% better absolute-value tolerance and matching than their diffused counterparts, which depend on the diffusion of dopants into the semiconductor surface (Fig. A).

With ion implantation, p-channel JFETs can be built on the same semiconductor surface as standard bipolar components. JFETs are basically implanted resistors with a second n-implant on top of the channel region that serves as a gate (Fig. B). They are useful in high-input-impedance amplifiers, automatic gaincontrol circuits, current sources, analog switches, multiplexers, and similar applications.

**Double isolation.** Double isolation, also called "iso-up," employs a rich ion-implanted p-type deposit in the substrate prior to growing the epitaxial layer. With conventional top-bottom diffusion, the bottom deposit diffuses downward. If everything is aligned correctly, the two deposits meet in the middle (Fig. C), leading not only to substantial reductions in lateral diffusions, but also to a smaller die.

Contrasting with that iso-up process is the traditional single isolation, which depends on a rich p-type diffusion penetrating through the epitaxial layer and into the substrate, thereby isolating regions of the epitaxial layer (Fig. D). This method does have its disadvantages, though. For one, it requires a long, high-temperature diffusion cycle. More critical to the user, however are the wider spacing needed between the isolation and subsequent diffusions to permit the significant lateral diffusion of the isolation. That extra spacing accumulates across the die, consuming a large amount of silicon. With the cost of manufacturing the die related largely to its area, the wasted space is reflected in a higher selling price to the user.

**Oxide capacitance.** By adding a special oxidation step to the processing, oxide capacitors are easily placed on a semicustom IC (Fig. E). These com-





ponents are identical in construction to those that internally compensate standard op amps; but unlike junction capacitors, they are not voltage-dependent.

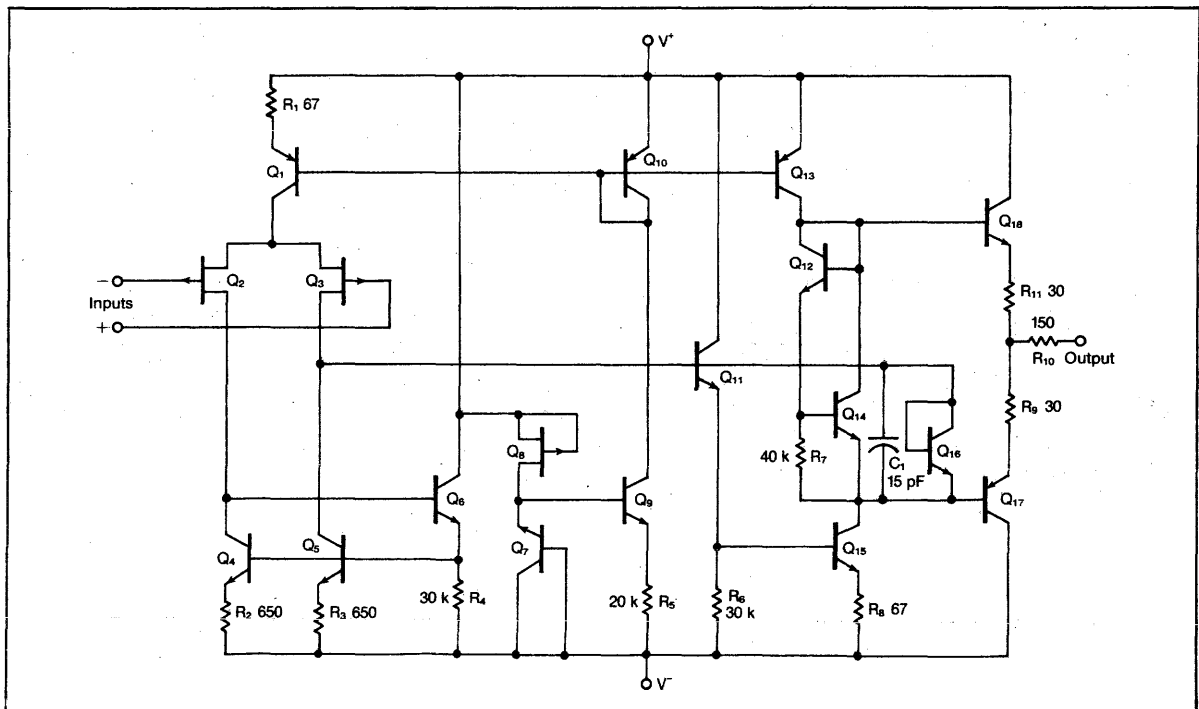
**36-V process.** In the past, high-density semicustom linear ICs were offered only with 20-V processing. Users needing 24V or the standard  $\pm 15$ -V op-amp supply were forced to externally regulate down to a voltage compatible with the chip or to place all high-voltage components off chip. Both cases severely limited higher levels of system integration. With the new 36-V process, these limitations become a thing of the past.

**Nitride passivation.** Applied over the standard purposes: it helps to eliminate ionic contamination from the chip surface, and it seals the active chip area from moisture. Both functions are aimed at reducing failure rates and improving the chip's

overall reliability. Furthermore, with nitride passivation, a user can employ a plastic package in applications that might otherwise require a more expensive, hermetically sealed ceramic housing.

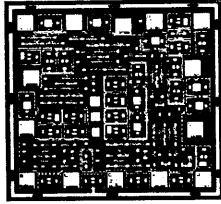
**IC components.** Though many characteristics of integrated components are similar to those of their discrete counterparts, there are important differences. Discrete components are individually produced and connected into circuits as required. Each component is manufactured by the most suitable process, thus optimizing performance. In the microscopic setting of integrated circuitry, such luxuries do not exist. Because the components are produced simultaneously, compromises must be made on performance. Despite these limitations, integrated components have unique advantages that allow them to overcome most design problems.

## Master Chip System typical building block cell

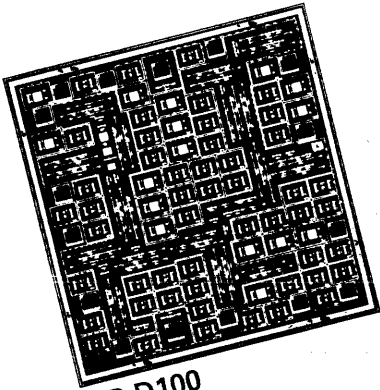


**A typical building-block cell in the Master Chip system, this JFET operational amplifier is similar to one of Exar's standard discrete op amps. This type of high-performance device relieves designers of the difficult task of creating their own amplifiers to satisfy each new application.**

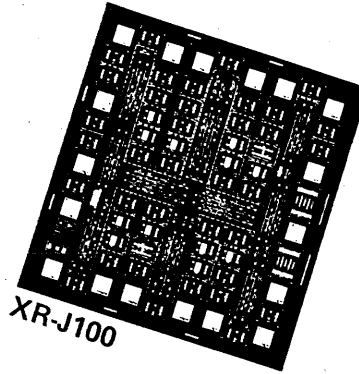
# Semi-Custom Arrays



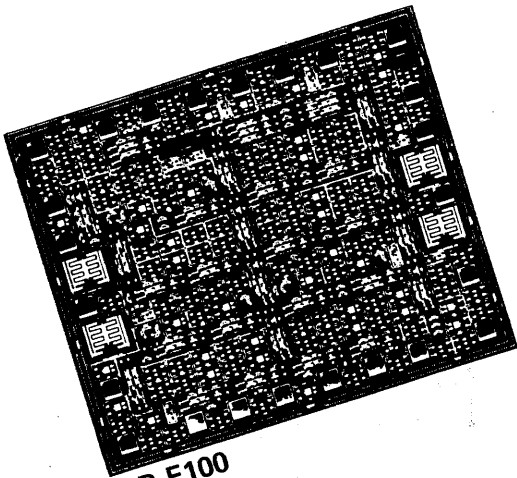
XR-C100A



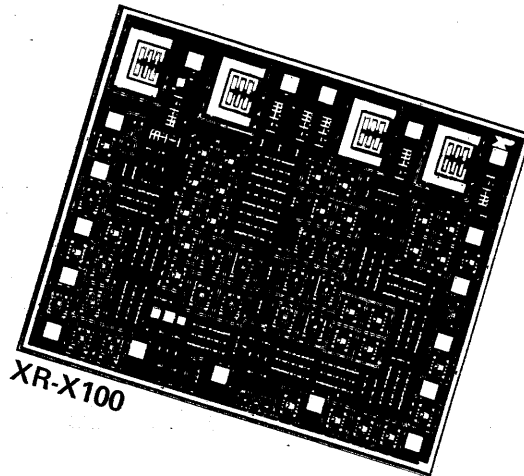
XR-D100



XR-J100



XR-F100



XR-X100

LINEAR

Exar Integrated Systems

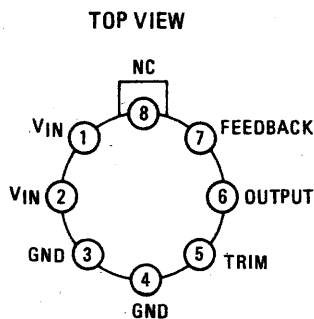
### FEATURES

- MONOLITHIC CONSTRUCTION
- INITIAL ACCURACY  $+10V \pm 0.010V$
- OUTPUT VOLTAGE ERROR, TOTAL  $\pm 1/4$  LSB
- LOW NOISE  $20\mu V_{p-p}$
- WIDE INPUT RANGE 12V TO 30V
- LOW POWER DISSIPATION 30mW
- OUTPUT SHORT CIRCUIT PROTECTION
- ADJUSTABLE OUTPUT

### APPLICATIONS

- AN ECONOMICAL EXTERNAL REFERENCE FOR: HI-5608; DAC 08; AD1408; AD559
- VOLTAGE REGULATOR REFERENCE
- PORTABLE BATTERY OPERATED EQUIPMENT
- NEGATIVE 10V REFERENCE

### PINOUT



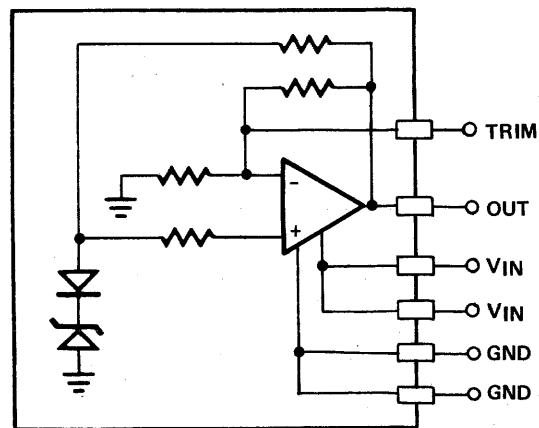
### DESCRIPTION

HA-1608 is a monolithic +10V adjustable voltage reference featuring accuracy and temperature stability specifications detailed exclusively for 8 bit data conversion systems. A stable +10V output is provided by a reference zener and buffer amplifier coupled with laser trimmed feedback and zener bias resistors. Long term stability is ensured through integration of all reference components into a monolithic design. Flexibility of HA-1608 is provided through an external trim control which allows the user to adjust the output voltage for binary or BCD applications without affecting overall performance.

These devices provide a total output voltage error of  $\pm 1/4$  LSB for 8 bit D/A or A/D converters. Low standby power (0.3mW) makes HA-1608 a natural selection for portable battery operated equipment, comparator references, and reference stacking circuits. These devices can also be used on -10V references.

HA-1608 is packaged in 8 pin metal cans (TO-99) and the pinout is arranged for convenient replacement of other less accurate regulators in applications demanding minimal change with temperature and time. HA-1608-2 is specified for -55°C to +125°C operation while the HA-1608-5 operates from 0°C to +75°C.

### FUNCTIONAL SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Input Voltage	40V	Operating Temperature Range	
Output Short Circuit Duration	Indefinitely	HA-1608-2	-55°C to +125°C
Power Dissipation	500mW	HA-1608-5	0°C to +75°C
Storage Temperature Range	-65°C to +150°C		

## ELECTRICAL CHARACTERISTICS (Note 2) ( $V_{IN} = +15V$ , $I_L = 0mA$ , unless otherwise specified)

PARAMETER	TEMP	HA-1608-2 -55°C to +125°C			HA-1608-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>POWER INPUT CHARACTERISTICS</b>								
Input Voltage Range, $V_{IN}$	Full	12	15	30	12	15	30	V
Quiescent Current, $I_Q$	25°C		1.9			1.9		mA
	Full			3.0			3.0	
<b>REGULATED OUTPUT CHARA.'S</b>								
Output Voltage, $V_O$	25°C	9.990	10.00	10.010	9.990	10.00	10.010	V
Output Load Current, $I_L$	Full	10	20		10	20		mA
Line Regulation ( $V_{IN} = 12V$ to $30V$ )	25°C		0.006			0.006		%V
	Full			0.015			0.015	
Load Regulation ( $I_L =$ Open to $10mA$ )	25°C		0.006			0.006		%mA
	Full			0.015			0.015	
Output Voltage Error Total $I_L = 0mA$ (Relative to 8-bit accuracy, see Definition # 3)	Full			±1/4 LSB			±1/4 LSB	
Output Noise Voltage, $E_N$ 0.1Hz to 10Hz	Full		35			35		$\mu V_{p-p}$
Dynamic Load Settling Time to ±0.1% to ±0.01%	25°C		2.5			2.5		$\mu s$
	25°C		5			5		
Warm-up Time (to ±0.01%)	25°C		1			1		sec
	Full		3			3		

### NOTES:

1. Absolute maximum ratings are limiting values beyond which the serviceability of the circuit may be impaired. Functional operation under any of these conditions is not necessarily implied.
2. The specified electrical characteristics apply to suggested hook-up only.



# HARRIS

# HA-2400/2404/2405

## PRAM Four Channel Programmable Amplifier

### FEATURES

- PROGRAMMABILITY
- HIGH SLEW RATE 30V/ $\mu$ s
- WIDE GAIN BANDWIDTH 40MHz
- HIGH GAIN 150,000
- LOW OFFSET CURRENT 5nA
- HIGH INPUT IMPEDANCE 30M  $\Omega$
- SINGLE CAPACITOR COMPENSATION
- DTL/TTL COMPATIBLE INPUTS

### DESCRIPTION

HA-2400/2404/2405 comprise a series of four-channel programmable amplifiers providing a level of versatility unsurpassed by any other monolithic operational amplifier. Versatility is achieved by employing four input amplifier channels, any one (or none) of which may be electronically selected and connected to a single output stage through DTL/TTL compatible address inputs. The device formed by the output and the selected pair of inputs is an op amp which delivers excellent slew rate, gain bandwidth and power bandwidth performance. Other advantageous features for these dielectrically isolated amplifiers include high voltage gain and input impedance coupled with low input offset voltage and offset current. External compensation is not required on this device at closed loop gains greater than 10.

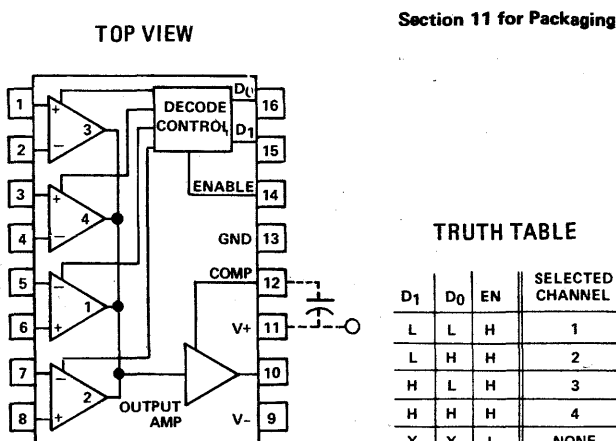
### APPLICATIONS

- THOUSANDS OF NEW APPLICATIONS; PROGRAM
  - SIGNAL SELECTION/MULTIPLEXING
  - OP AMP GAIN
  - OSCILLATOR FREQUENCY
  - FILTER CHARACTERISTICS
  - ADD-SUBTRACT FUNCTIONS
  - INTEGRATOR CHARACTERISTICS
  - COMPARATOR LEVELS

Each channel of the HA-2400/2404/2405 can be controlled and operated with suitable feedback networks in any of the standard op amp configurations. This specialization makes these amplifiers excellent components for multiplexing, signal selection, and mathematical function designs. With 30V/ $\mu$ s slew rate, 40MHz gain bandwidth, and 30M ohms input impedance these devices are ideal building blocks for signal generators, active filters, and data acquisition designs. Programmability coupled with 2mV typical offset voltage and 5nA offset current makes these amplifiers outstanding components for signal conditioning circuits.

HA-2400/2404/2405 are available in a 16 pin dual-in-line package. HA-2400 is specified from -55°C to +125°C. HA-2404 is specified over the -25°C to +85°C range, while HA-2405 operates from 0°C to +75°C.

### PINOUT



### SCHEMATIC

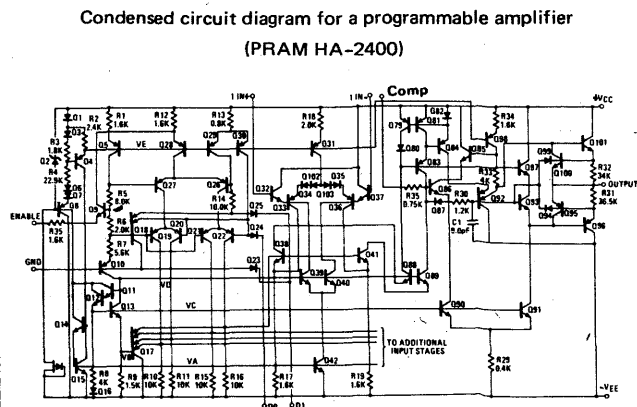


Diagram includes: ONE INPUT STAGE, DECODE CONTROL, BIAS NETWORK AND OUTPUT STAGE

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V- Terminals	45.0V	Internal Power Dissipation (Note 13)	300mW
Differential Input Voltage	$\pm V_{Supply}$	Operating Temperature Range	$-55^{\circ}C \leq T_A \leq +125^{\circ}C$ (HA-2400) $-25^{\circ}C \leq T_A \leq +85^{\circ}C$ (HA-2404) $0^{\circ}C \leq T_A \leq +75^{\circ}C$ (HA-2405)
Digital Input Voltage	-0.76V to +10.0V	Storage Temperature Range	$-65^{\circ}C \leq T_A \leq +150^{\circ}C$
Output Current	Short Circuit Protected ( $I_{SC} \leq \pm 33mA$ )		

## ELECTRICAL CHARACTERISTICS

Test Conditions:  $V_{Supply} = \pm 15.0V$  unless otherwise specified.

Digital inputs:  $V_{IL} = +0.5V$ ,  $V_{IH} = +2.4V$   
Limits apply to each of the four channels, when addressed.

PARAMETER	TEMP.	HA-2400/HA-2404 LIMITS			HA-2405 LIMITS			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C Full		4 11	9 11		4 11	9 11	mV mV
Bias Current (Note 12)	+25°C Full		50 400	200 400		50 500	250 500	nA nA
Offset Current (Note 12)	+25°C Full		5 100	50 100		5 100	50 100	nA nA
Input Resistance (Note 12)	+25°C		30			30		MΩ
Common Mode Range	Full	$\pm 9.0$			$\pm 9.0$			V
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 1,5)	+25°C Full	50K 25K	150K		50K 25K	150K		V/V V/V
Common Mode Rejection Ratio (Note 2)	Full	80	100		74	100		dB
Gain Bandwidth (Note 3) (Note 4)	+25°C +25°C	20 4	40 8		20 4	40 8		MHz MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 1)	Full	$\pm 10.0$	$\pm 12.0$		$\pm 10.0$	$\pm 12.0$		V
Output Current	+25°C	10	20		10	20		mA
Full Power Bandwidth (Notes 3, 5) (Notes 4,5)	+25°C +25°C	200 100	500 200		200 100	500 200		kHz kHz
<b>TRANSIENT RESPONSE</b>								
Rise Time (Notes 4,6)	+25°C		20	45		20	50	ns
Overshoot (Notes 4,6)	+25°C		25	40		25	40	%
Slew Rate (Notes 3,7) (Notes 4,7)	+25°C +25°C	20 6	30 8		20 6	30 8		V/μs V/μs
Settling Time (Notes 4, 7, 8)	+25°C		1.5	2.5		1.5	2.5	μs
<b>CHANNEL SELECT CHARACTERISTICS</b>								
Digital Input Current ( $V_{IN} = 0V$ )	Full		1	1.5		1	1.5	mA
Digital Input Current ( $V_{IN} = +5.0V$ )	Full		5			5		nA
Output Delay (Note 9)	+25°C		100	250		100	250	ns
Crosstalk (Note 10)	+25°C	-80	-110		-74	-110		dB
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	+25°C		4.8	6.0		4.8	6.0	mA
Power Supply Rejection Ratio (Note 11)	Full	74	90		74	90		dB

- NOTES: 1.  $R_L = 2K\Omega$   
2.  $V_{CM} = \pm 5V.D.C.$   
3.  $A_V = +10$ ,  $C_{COMP} = 0$ ,  $R_L = 2K\Omega$ ,  $C_L = 50pF$   
4.  $A_V = +1$ ,  $C_{COMP} = 15pF$ ,  $R_L = 2K\Omega$ ,  $C_L = 50pF$   
5.  $V_{OUT} = 20V$  peak-to-peak  
6.  $V_{OUT} = 200mV$  peak-to-peak  
7.  $V_{OUT} = 10.0V$  peak-to-peak

8. To 0.1% of final value  
9. To 10% of final value; output then slews at normal rate to final value.  
10. Unselected input to output;  $V_{IN} = \pm 10 V.D.C.$   
11.  $V_{SUPP} = \pm 10V.D.C.$  to  $\pm 20V.D.C.$   
12. Unselected channels have approximately the same input parameters.  
13. Derate by  $4.3mW/^{\circ}C$  above  $105^{\circ}C$

LINEAR

Harris Semiconductor

### FEATURES

- HIGH SLEW RATE 30V/ $\mu$ S
- FAST SETTLING 330ns
- WIDE POWER BANDWIDTH 500kHz
- HIGH GAIN BANDWIDTH 12MHz
- HIGH INPUT IMPEDANCE 50 M $\Omega$
- LOW OFFSET CURRENT 10nA
- INTERNALLY COMPENSATED

### APPLICATIONS

- DATA ACQUISITION SYSTEMS
- R.F. AMPLIFIERS
- VIDEO AMPLIFIERS
- SIGNAL GENERATORS
- PULSE AMPLIFICATION

### DESCRIPTION

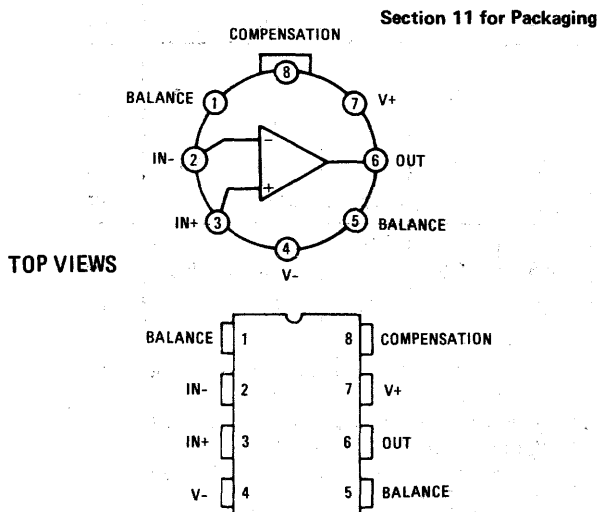
HA-2500/2502/2505 comprise a series of monolithic operational amplifiers whose designs are optimized to deliver excellent slew rate, bandwidth, and settling time specifications. The outstanding dynamic features of this internally compensated device are complemented with low offset voltage and offset current.

These dielectrically isolated amplifiers are ideally suited for applications such as data acquisition, R.F., video, and pulse conditioning circuits. Slew rate of  $\pm 25V/\mu s$  and 330ns (0.1%) settling time make these devices excellent components in fast, accurate data acquisition and pulse amplification designs. 12 MHz bandwidth and 500kHz power bandwidth make these devices well suited to R.F. and video applications. With 2mV typical offset voltage plus offset trim capability and 10nA offset current, HA-2500/2502/2505 are particularly useful components in signal conditioning designs.

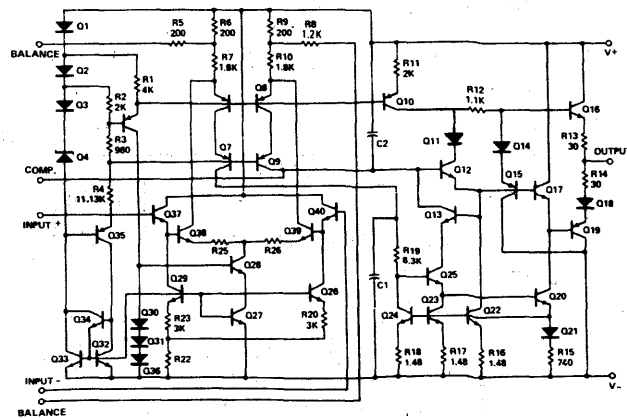
The gain and offset voltage figures of the HA-2500 series are optimized by internal component value changes while the similar design of the HA-2510 series is maximized for slew rate.

HA-2500/2502/2505 are available in metal can (TO-99) packages. HA-2500 and HA-2502 are specified over the  $-55^{\circ}C$  to  $+125^{\circ}C$  range. HA-2505 is specified from  $0^{\circ}C$  to  $+75^{\circ}C$ .

### PINOUT



### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V<sup>+</sup> and V<sup>-</sup> Terminals 40.0V  
 Differential Input Voltage ±15.0V  
 Peak Output Current 50mA  
 Internal Power Dissipation 300mW

Operating Temperature Range – HA-2500/HA-2502  
 HA-2505  
 Storage Temperature Range

-55°C ≤ T<sub>A</sub> ≤ +125°C  
 0°C ≤ T<sub>A</sub> ≤ +75°C  
 -65°C ≤ T<sub>A</sub> ≤ +150°C

## ELECTRICAL CHARACTERISTICS

V<sup>+</sup> = +15V D.C., V<sup>-</sup> = -15V D.C.

PARAMETER	TEMP.	HA-2500 -55°C to +125°C			HA-2502 -55°C to +125°C			HA-2505 0°C to +75°C			UNITS
		LIMITS			LIMITS			LIMITS			
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C Full		2 5 8		4 8 10		4 8 10		4 8 10		mV mV
Offset Voltage Average Drift	Full		20		20		20		20		μV/°C
Bias Current	+25°C Full		100 200 400		125 250 500		125 250 500		125 250 500		nA nA
Offset Current	+25°C Full		10 25 50		20 50 100		20 50 100		20 50 100		nA nA
Input Resistance (Note 10)	+25°C	25	50		20	50		20	50		MΩ
Common Mode Range	Full	±10.0			±10.0			±10.0			V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Note 1,4)	+25°C Full	20K 15K	30K		15K 10K	25K		15K 10K	25K		V/V V/V
Common Mode Rejection Ratio (Note 2)	Full	80	90		74	90		74	90		dB
Gain Bandwidth Product (Note 3)	+25°C		12			12			12		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 1)	Full	±10.0	±12.0		±10.0	±12.0		±10.0	±12.0		V
Output Current (Note 4)	+25°C	±10	±20		±10	±20		±10	±20		mA
Full Power Bandwidth (Note 4)	+25°C	350	500		300	500		300	500		kHz
<b>TRANSIENT RESPONSE</b>											
Rise Time (Notes 1, 5, 6 & 8)	+25°C		25 50		25 50		25 50		25 50		ns
Overshoot (Notes 1, 5, 7 & 8)	+25°C		25 40		25 50		25 50		25 50		%
Slew Rate (Notes 1, 5, 8 & 12)	+25°C	±25	±30		±20	±30		±20	±30		V/μs
Settling Time to 0.1% (Notes 1, 5, 8 & 12)	+25°C		0.33			0.33			0.33		μs
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	+25°C		4 6		4 6		4 6		4 6		mA
Power Supply Rejection Ratio (Note 9)	Full	80	90		74	90		74	90		dB

- NOTES: 1. R<sub>L</sub> = 2K  
 2. V<sub>CM</sub> = ±10V  
 3. A<sub>V</sub> > 10  
 4. V<sub>O</sub> = ±10.0V  
 5. C<sub>L</sub> = 50pF  
 6. V<sub>O</sub> = ±200mV

7. V<sub>O</sub> ± 200mV  
 8. See transient response test circuits and waveforms page four.  
 9. ΔV = ±5.0V

10. This parameter value is based on design calculations.  
 11. Full power bandwidth guaranteed based on slew rate measurement using FPBW = S.R./2π V<sub>peak</sub>.  
 12. V<sub>OUT</sub> = ±5V

LINEAR

Harris Semiconductor



# HA-2510/2512/2515

## High Slew Rate Operational Amplifiers

### FEATURES

- HIGH SLEW RATE 60V/ $\mu$ s
- FAST SETTLING 250ns
- WIDE POWER BANDWIDTH 1,000kHz
- HIGH GAIN BANDWIDTH 12MHz
- HIGH INPUT IMPEDANCE 100M $\Omega$
- LOW OFFSET CURRENT 10nA
- INTERNALLY COMPENSATED

### APPLICATIONS

- DATA ACQUISITION SYSTEMS
- R.F. AMPLIFIERS
- VIDEO AMPLIFIERS
- SIGNAL GENERATORS
- PULSE AMPLIFICATION

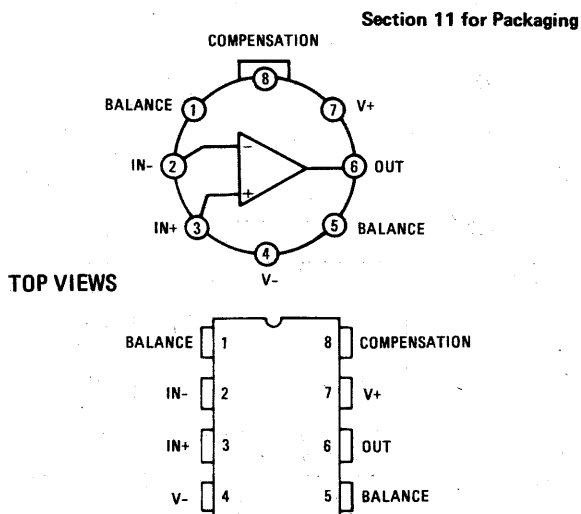
### DESCRIPTION

The HA-2510/2512/2515 are a series of high performance operational amplifiers which set the standards for maximum slew rate, highest accuracy and widest bandwidth for internally compensated monolithic devices. In addition to excellent dynamic characteristics, these dielectrically isolated amplifiers also offer low offset current and high input impedance.

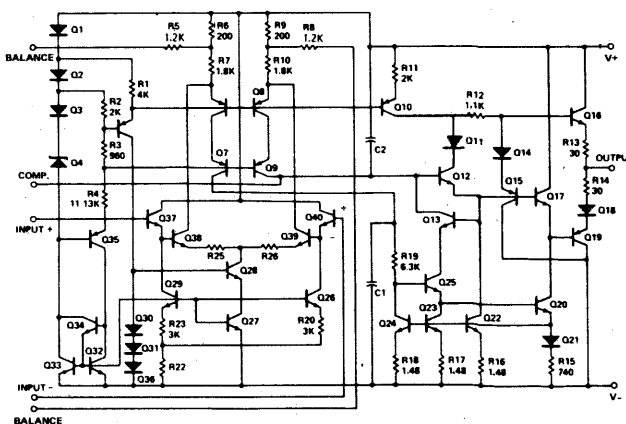
The  $\pm 60V/\mu s$  slew rate and 250ns (0.1%) settling time of these amplifiers is ideally suited for high speed D/A, A/D, and pulse amplification designs. HA-2510/2512/2515's superior 12MHz gain bandwidth and 1000kHz power bandwidth is extremely useful in R.F. and video applications. For accurate signal conditioning these amplifiers also provide 10nA offset current, coupled with 100M $\Omega$  input impedance, and offset trim capability.

The HA-2510/2512 are available in metal can (TO-99) and 14-pin flat packages. HA-2510 and HA-2512 are specified from -55 $^{\circ}$ C to +125 $^{\circ}$ C. HA-2515 is specified over the 0 $^{\circ}$ C to +75 $^{\circ}$ C range, and is available in the TO-99 package.

### PINOUT



### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V <sup>+</sup> and V <sup>-</sup> Terminals	40.0V	Peak Output Current	50mA
Differential Input Voltage	±15.0V	Internal Power Dissipation	300mW
Operating Temperature Range		Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C
HA-2510/HA-2512	-55°C ≤ T <sub>A</sub> ≤ +125°C		
HA-2515	0°C ≤ T <sub>A</sub> ≤ +75°C		

## ELECTRICAL CHARACTERISTICS

V<sup>+</sup> = +15V D.C., V<sup>-</sup> = 15V D.C.

PARAMETER	TEMP.	HA-2510 -55°C to +125°C			HA-2512 -55°C to +125°C			HA-2515 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C Full		4 8	11		5 10	14		5 10	14	mV mV
Offset Voltage Average Drift	Full		20			25			30		μV/°C
Bias Current	+25°C Full		100 200	400		125 250	500		125 250	500	nA nA
Offset Current	+25°C Full		10 25	50		20 50	100		20 50	100	nA nA
Input Resistance (Note 10)	+25°C	50	100		40	100		40	100		MΩ
Common Mode Range	Full	±10.0			±10.0			±10.0			V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Note 1,4)	+25°C Full	10K 7.5K	15K		7.5K 5K	15K		7.5K 5K	15K		V/V V/V
Common Mode Rejection Ratio (Note 2)	Full	80	90		74	90		74	90		dB
Gain Bandwidth Product (Note 3)	+25°C		12			12			12		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 1)	Full	±10.0	±12.0		±10.0	±12.0		±10.0	±12.0		V
Output Current (Note 4)	+25°C	±10	±20		±10	±20		±10	±20		mA
Full Power Bandwidth (Note 4, 11)	+25°C	750	1000		600	1000		600	1000		kHz
<b>TRANSIENT RESPONSE</b>											
Rise Time (Notes 1, 5, 6 & 8)	+25°C		25	50		25	50		25	50	ns
Overshoot (Notes 1, 5, 7 & 8)	+25°C		25	40		25	50		25	50	%
Slew Rate (Notes 1, 5, 8 & 12)	+25°C	±50	±65		±40	±60		±40	±60		V/μs
Settling Time (Notes 1, 5, 8 & 12)	+25°C		0.25			0.25			0.25		μs
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	+25°C		4	6		4	6		4	6	mA
Power Supply Rejection Ratio (Note 9)	Full	80	90		74	90		74	90		dB

- NOTES:
- R<sub>L</sub> = 2K
  - V<sub>CM</sub> = ±10V
  - A<sub>V</sub> > 10
  - V<sub>O</sub> = ±10.0V
  - C<sub>L</sub> = 50pF
  - V<sub>O</sub> = ±200mV
  - V<sub>O</sub> = ±200mV
  - See transient response test circuits and waveforms
  - ΔV = ±5.0V
  - This parameter value is based upon design calculations.
  - Full power bandwidth guaranteed based upon slew rate measurement  
FPBW = S.R./2πV<sub>peak</sub>.
  - V<sub>OUT</sub> = ±5V

## Uncompensated High Slew Rate Operational Amplifiers

### FEATURES

- |                        |               |
|------------------------|---------------|
| • HIGH SLEW RATE       | 120V/ $\mu$ s |
| • FAST SETTLING        | 200ns         |
| • WIDE POWER BANDWIDTH | 2,000kHz      |
| • HIGH GAIN BANDWIDTH  | 20MHz         |
| • HIGH INPUT IMPEDANCE | 100M $\Omega$ |
| • LOW OFFSET CURRENT   | 10nA          |

### DESCRIPTION

HA-2520/2522/2525 comprise a series of monolithic operational amplifiers delivering an unsurpassed combination of specifications for slew rate, bandwidth and settling time. These dielectrically isolated amplifiers are controlled at closed loop gains greater than 3 without external compensation. In addition, these high performance components also provide low offset current and high input impedance.

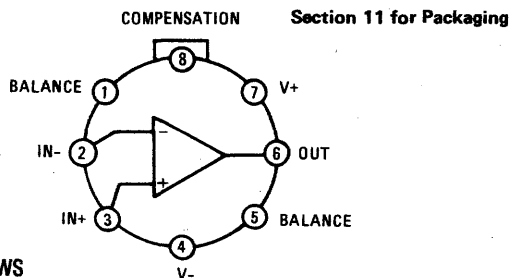
120V/ $\mu$ s slew rate and 200ns (0.1%) settling time of these amplifiers make them ideal components for pulse amplification and data acquisition designs. These devices are valuable components for R.F. and video circuitry requiring up to 20MHz gain bandwidth and 2MHz power bandwidth. For accurate signal conditioning designs the HA-2520/2522/2525's superior dynamic specifications are complimented by 10nA offset current, 100M $\Omega$  input impedance and offset trim capability.

The HA-2520/2522 are available in metal can (TO-99) and 14-pin flat packages. HA-2520 and HA-2522 are specified over -55 $^{\circ}$ C to +125 $^{\circ}$ C range. HA-2525 is specified from 0 $^{\circ}$ C to +75 $^{\circ}$ C, and is available in the TO-99 package.

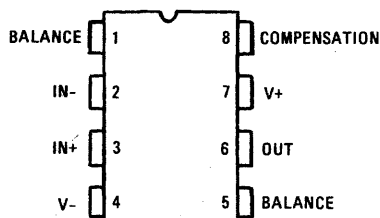
### APPLICATIONS

- DATA ACQUISITION SYSTEMS
- R.F. AMPLIFIERS
- VIDEO AMPLIFIERS
- SIGNAL GENERATORS
- PULSE AMPLIFICATION

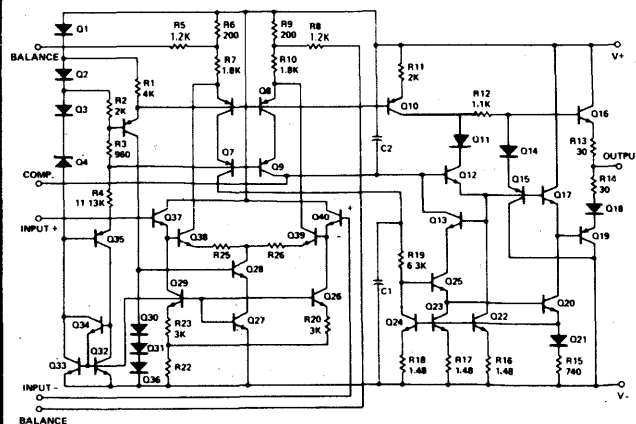
### PINOUT



### TOP VIEWS



### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V <sup>+</sup> and V <sup>-</sup> Terminals	40.0V	Peak Output Current	50mA
Differential Input Voltage	±15.0V	Internal Power Dissipation	300mW
Operating Temperature Range		Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C
HA-2520/2522	-55°C ≤ T <sub>A</sub> ≤ +125°C		
HA-2525	0°C ≤ T <sub>A</sub> ≤ +75°C		

## ELECTRICAL CHARACTERISTICS

V<sup>+</sup> = +15V D.C., V<sup>-</sup> = -15V D.C.

PARAMETER	TEMP.	HA-2520 -55°C to +125°C			HA-2522 -55°C to +125°C			HA-2525 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C Full		4 8 11		5 10 14		5 10 14				mV mV
Offset Voltage Average Drift	Full		20		25		30				μV/°C
Bias Current	+25°C Full		100 200 400		125 250 500		125 250 500				nA nA
Offset Current	+25°C Full		10 25 50		20 50 100		20 50 100				nA nA
Input Resistance (Note 9)	+25°C	50	100		40	100		40	100		MΩ
Common Mode Range	Full	±10.0			±10.0			±10.0			V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Note 1,4)	+25°C Full	10K 7.5K	15K		7.5K 5K	15K		7.5K 5K	15K		V/V V/V
Common Mode Rejection Ratio (Note 2)	Full	80	90		74	90		74	90		dB
Gain Bandwidth Product (Note 3)	+25°C	10	20		10	20		10	20		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 1)	Full	±10.0	±12.0		±10.0	±12.0		±10.0	±12.0		V
Output Current (Note 4)	+25°C	±10	±20		±10	±20		±10	±20		mA
Full Power Bandwidth (Note 4, 10)	+25°C	1500	2000		1200	1600		1200	1600		kHz
<b>TRANSIENT RESPONSE (A<sub>V</sub> = +3)</b>											
Rise Time (Notes 1, 5, 6 & 8)	+25°C		25 50		25 50		25 50		25 50		ns
Overshoot (Notes 1, 5, 6 & 8)	+25°C		25 40		25 50		25 50		25 50		%
Slew Rate (Notes 1, 5, 8 & 11)	+25°C	±100	±120		±80	±120		±80	±120		V/μs
Settling Time (Notes 1, 5, 8 & 11)	+25°C		0.20		0.20		0.20		0.20		μs
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	+25°C		4 6		4 6		4 6		4 6		mA
Power Supply Rejection Ratio (Note 7)	Full	80	90		74	90		74	90		dB

- NOTES: 1. R<sub>L</sub> = 2K  
 2. V<sub>CM</sub> = ±10V  
 3. A<sub>V</sub> > 10  
 4. V<sub>O</sub> = ±10.0V  
 5. C<sub>L</sub> = 50pF  
 6. V<sub>O</sub> = ±200mV  
 7. ΔV = ±5.0V  
 8. See transient response test circuits and waveforms  
 9. This parameter value is based upon design calculations.  
 10. Full power bandwidth guaranteed based upon slew rate measurement  
 FPBW = S.R./2πV<sub>peak</sub>.  
 11. V<sub>OUT</sub> = ±5V

# HA-2539

*Very High Slew Rate  
Wideband  
Operational Amplifiers*

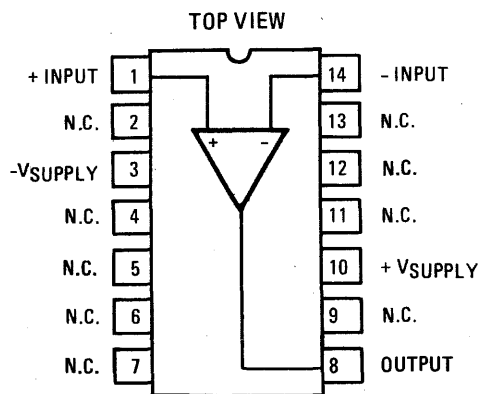
## FEATURES

- VERY HIGH SLEW RATE 600V/ $\mu$ s
- OPEN LOOP GAIN 30kV/V
- WIDE GAIN-BANDWIDTH 600MHz
- POWER BANDWIDTH 9.5MHz
- LOW OFFSET VOLTAGE 3mV
- INPUT VOLTAGE NOISE 15nV/ $\sqrt{\text{Hz}}$
- OUTPUT VOLTAGE SWING  $\pm 10\text{V}$

## APPLICATIONS

- PULSE AND VIDEO AMPLIFIERS
- WIDEBAND AMPLIFIERS
- HIGH SPEED SAMPLE-HOLD CIRCUITS
- RF OSCILLATORS

## PINOUT



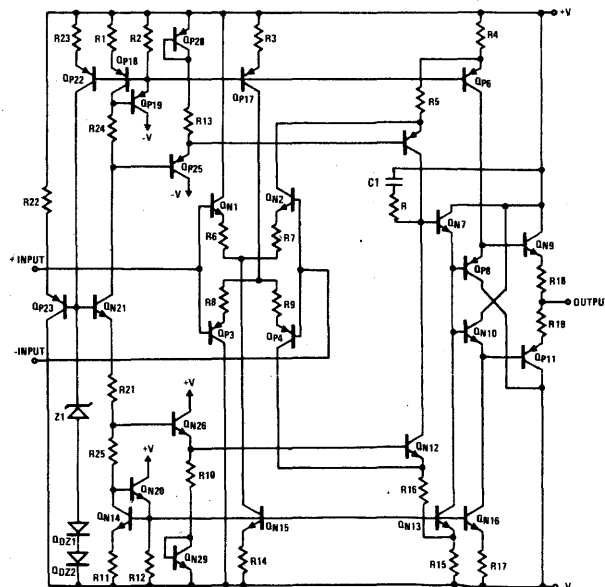
## GENERAL DESCRIPTION

The Harris HA-2539 represents the ultimate in high slew rate wideband, monolithic, operational amplifiers. It has been designed and constructed with the Harris high frequency BIPDIP (Bipolar dielectric isolation process), and features dynamic parameters never before available from a truly differential device.

With a 600V/ $\mu$ s slew rate and a 600MHz gain-bandwidth-product, the HA-2539 is ideally suited for use in video and RF amplifier designs. Full  $\pm 10\text{V}$  output swing coupled with outstanding A.C. parameters and complemented by high open loop gain makes these devices useful in high speed data acquisition systems.

The HA-2539 is available in the 14 pin CERDIP. The HA-2539-2 denotes  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  operation while the HA-2539-5 operates over the  $0^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  range.

## SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Voltage between V+ and V- Terminals	35V
Differential Input Voltage	6V
Output Current	50mA (Peak)
Internal Power Dissipation (Note 2)	870mW (Cerdip)
Operating Temperature Range: (HA-2539-2)	$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
(HA-2539-5)	$0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$
Storage Temperature Range	$-65^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$

## ELECTRICAL CHARACTERISTICS $V_{\text{SUPPLY}} = \pm 15$ Volts; $R_L = 1\text{K}$ ohms, unless otherwise specified.

PARAMETER	TEMP	HA-2539-2 -55°C to +125°C			HA-2539-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		3	5		3	15	mV
	FULL			10			20	mV
Average Offset Voltage Drift	FULL		20			20		$\mu\text{V}/^{\circ}\text{C}$
Bias Current	+25°C		5	20		5	20	$\mu\text{A}$
	FULL			25			25	$\mu\text{A}$
Offset Current	+25°C		1	6		1	6	$\mu\text{A}$
	FULL			8			8	$\mu\text{A}$
Input Resistance	+25°C		10			10		Kohms
Input Capacitance	+25°C		1.0			1.0		pF
Common Mode Range	FULL	$\pm 10$			$\pm 10$			V
Input Voltage Noise (f = 1kHz, $R_g = 0\Omega$ )	+25°C		15			15		$\text{nV}/\sqrt{\text{Hz}}$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 3)	+25°C	15K	30K		10K	30K		V/V
	FULL	5K			5K			V/V
Common-Mode Rejection Ratio (Note 4)	FULL	60			60			dB
Gain-Bandwidth-Product (Notes 5 & 6)	+25°C		600			600		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 3)	FULL	$\pm 10$			$\pm 10$			V
Output Current (Note 3)	+25°C	10			10			mA
Output Resistance	+25°C		30			30		Ohms
Full Power Bandwidth (Note 3 & 7)	+25°C	8.7	9.5		8.7	9.5		MHz
<b>TRANSIENT RESPONSE (Note 8)</b>								
Rise Time	+25°C		7			7		ns
Overshoot	+25°C		15			8		%
Slew Rate	+25°C	550	600		550	600		$\text{V}/\mu\text{s}$
Settling Time: 10V Step to 0.1%	+25°C		350			350		ns
<b>POWER REQUIREMENTS</b>								
Supply Current	FULL		20	25		20	25	mA
Power Supply Rejection Ratio (Note 9)	FULL	60			60			dB

LINEAR  
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### FEATURES

- VERY HIGH SLEW RATE 400V/ $\mu$ s
- FAST SETTling TIME 250ns
- WIDE GAIN-BANDWIDTH 400MHz
- POWER BANDWIDTH 6MHz
- LOW OFFSET VOLTAGE 5mV
- INPUT VOLTAGE NOISE 15nV/ $\sqrt{\text{Hz}}$
- OUTPUT VOLTAGE SWING  $\pm 10\text{V}$
- MONOLITHIC BIPOLAR CONSTRUCTION

### APPLICATIONS

- PULSE AND VIDEO AMPLIFIERS
- WIDEBAND AMPLIFIERS
- HIGH SPEED SAMPLE-HOLD CIRCUITS
- FAST, PRECISE D/A CONVERTERS

### GENERAL DESCRIPTION

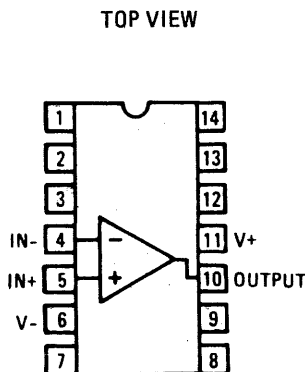
The Harris HA-2540 is a wideband, very high slew rate, monolithic operational amplifier featuring superior speed and bandwidth characteristics. Bipolar construction coupled with dielectric isolation allows this truly differential device to deliver outstanding performance. Additionally, the HA-2540 has a drive capability of  $\pm 10\text{V}$  into a 1K ohm load. Other desirable characteristics include low input voltage noise, low offset voltage, and fast settling time.

A 400V/ $\mu$ s slew rate ensures high performance in video and pulse amplification circuits, while the 400MHz gain-bandwidth-product is ideally suited for wideband signal amplification. A settling time of 250ns also makes the HA-2540 an excellent selection for high speed Data Acquisition Systems.

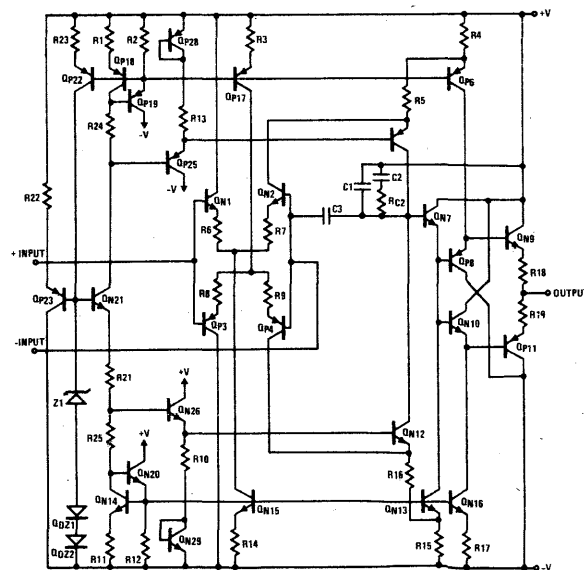
The HA-2540-2 is specified over the  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  range while the HA-2540-5 is specified from  $0^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$ .

### PINOUT

Section 11 for Packaging



### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Voltage between V+ and V- Terminals	35V
Differential Input Voltage	6V
Output Current	50mA (Peak)
Internal Power Dissipation (Note 2)	870mW (Cerdip)
Operating Temperature Range: (HA-2540-2)	$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
(HA-2540-5)	$0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$
Storage Temperature Range	$-65^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$

## ELECTRICAL CHARACTERISTICS $V_{\text{SUPPLY}} = \pm 15$ Volts; $R_L = 1\text{K}$ ohms, unless otherwise specified.

PARAMETER	TEMP	HA-2540-2 -55°C to +125°C			HA-2540-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		3	5		3	15	mV
	FULL			10			20	mV
Average Offset Voltage Drift	FULL		20			20		$\mu\text{V}/^{\circ}\text{C}$
Bias Current	+25°C		5	20		5	20	$\mu\text{A}$
	FULL			25			25	$\mu\text{A}$
Offset Current	+25°C		1	6		1	6	$\mu\text{A}$
	FULL			8			8	$\mu\text{A}$
Input Resistance	+25°C		10			10		Kohms
Input Capacitance	+25°C		1.0			1.0		pF
Common Mode Range	FULL	$\pm 10$			$\pm 10$			V
Input Noise Voltage ( $f = 1\text{kHz}$ , $R_g = 0\Omega$ )	+25°C		15			15		$\text{nV}/\sqrt{\text{Hz}}$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 3)	+25°C	15K	30K		10K	30K		V/V
	FULL	5K			5K			V/V
Common-Mode Rejection Ratio (Note 4)	FULL	60			60			dB
Gain-Bandwidth-Product (Notes 5 & 6)	+25°C		400			400		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 3)	FULL	$\pm 10$			$\pm 10$			V
Output Current (Note 3)	+25°C	10			10			mA
Output Resistance	+25°C		30			30		Ohms
Full Power Bandwidth (Note 3 & 7)	+25°C	5.5	6		5.5	6		MHz
<b>TRANSIENT RESPONSE (Note 8)</b>								
Rise Time	+25°C		14			14		ns
Overshoot	+25°C		5			5		%
Slew Rate	+25°C	350	400		350	400		V/ $\mu\text{s}$
Settling Time: 10V Step to 0.1%	+25°C		250			250		ns
<b>POWER REQUIREMENTS</b>								
Supply Current	FULL		20	25		20	25	mA
Power Supply Rejection Ratio (Note 9)	FULL	60			60			dB





# HARRIS

# HA-2600/2602/2605

## WideBand, High Impedance Operational Amplifiers

### FEATURES

- WIDE BANDWIDTH 12MHz
- HIGH INPUT IMPEDANCE 500MΩ
- LOW INPUT BIAS CURRENT 1nA
- LOW INPUT OFFSET CURRENT 1nA
- LOW INPUT OFFSET VOLTAGE 0.5mV
- HIGH GAIN 150K V/V
- HIGH SLEW RATE 7V/μs
- OUTPUT SHORT CIRCUIT PROTECTION

### APPLICATIONS

- VIDEO AMPLIFIER
- PULSE AMPLIFIER
- AUDIO AMPLIFIERS AND FILTERS
- HIGH-Q ACTIVE FILTERS
- HIGH-SPEED COMPARATORS
- LOW DISTORTION OSCILLATORS

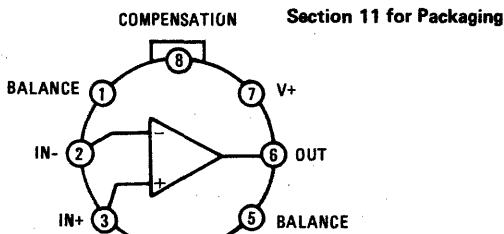
### DESCRIPTION

HA-2600/2602/2605 are internally compensated bipolar operational amplifiers that feature very high input impedance (500 MΩ, HA-2600) coupled with wideband AC performance. The high resistance of the input stage is complemented by low offset voltage (0.5mV, HA-2600) and low bias and offset current (1nA, HA-2600) to facilitate accurate signal processing. Input offset can be reduced further by means of an external nulling potentiometer. 12MHz unity gain-bandwidth product, 7V/μs slew rate and 150,000V/V open-loop gain enables HA-2600/2602/2605 to perform high-gain amplification of fast, wideband signals. These dynamic characteristics, coupled with fast settling times, make these amplifiers ideally suited to pulse amplification designs as well as high frequency (e.g. video) applications. The frequency response of the amplifier can be tailored to exact design requirements by means of an external bandwidth control capacitor.

In addition to its application in pulse and video amplifier designs, HA-2600/2602/2605 is particularly suited to other high performance designs such as high-gain low distortion audio amplifiers, high-Q and wideband active filters and high-speed comparators.

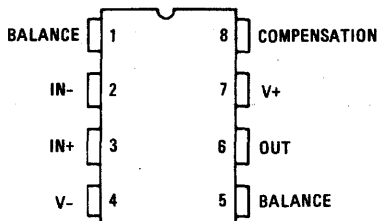
HA-2600 and HA-2602 are guaranteed over -55°C to +125°C. HA-2605 is specified from 0°C to +75°C. All devices are available in TO-99 cans, and HA-2600/2602 are available in 10 lead flat packages.

### PINOUT

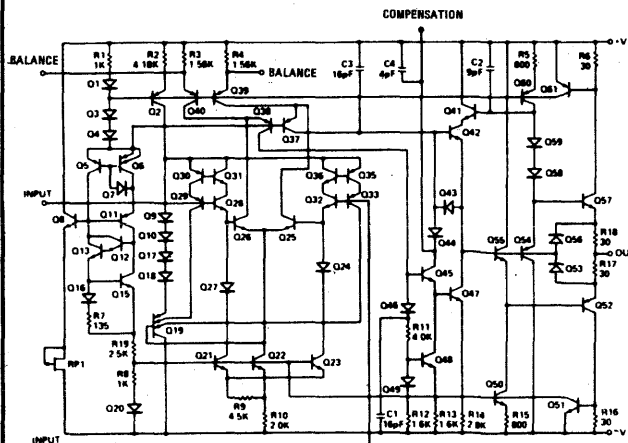


Case Connected to V-

TOP VIEWS



### SCHEMATIC



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# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V <sup>+</sup> and V <sup>-</sup> Terminals	45.0V
Differential Input Voltage	±12.0V
Peak Output Current	Full Short Circuit Protection
Internal Power Dissipation	300mW
Operating Temperature Range – HA-2600/HA-2602	-55°C ≤ T <sub>A</sub> ≤ +125°C
HA-2605	0° ≤ T <sub>A</sub> ≤ +75°C
Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C

## ELECTRICAL CHARACTERISTICS

V<sup>+</sup> = +15VDC, V<sup>-</sup> = -15VDC

PARAMETER	TEMP.	HA-2600 -55°C to +125°C			HA-2602 -55°C to +125°C			HA-2605 0°C to +75°C			UNITS
		LIMITS			LIMITS			LIMITS			
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C Full		0.5 2	4 6		3 7	5 7		3 7	5 7	mV mV
Offset Voltage Average Drift	Full		5								μV/°C
Bias Current	+25°C Full		1 10	10 30		15 25	25 60		5 5	25 40	nA nA
Offset Current	+25°C Full		1 5	10 30		5 25	25 60		5 5	25 40	nA nA
Input Resistance (Note 10)	+25°C	100	500		40	300		40	300		MΩ
Common Mode Range	Full	±11.0			±11.0			±11.0			V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Notes 1, 4)	+25°C Full	100K 70K	150K		80K 60K	150K		80K 70K	150K		V/V V/V
Common Mode Rejection Ratio (Note 2)	Full	80	100		74	100		74	100		dB
Unity Gain Bandwidth (Note 3)	+25°C		12			12			12		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 1)	Full	±10.0	±12.0		±10.0	±12.0		±10.0	±12.0		V
Output Current (Note 4)	+25°C	±15	±22		±10	±18		±10	±18		mA
Full Power Bandwidth (Note 4 & 11)	+25°C	50	75		50	75		50	75		kHz
<b>TRANSIENT RESPONSE</b>											
Rise Time (Notes 1, 5, 6 & 8)	+25°C		30	60		30	60		30	60	ns
Overshoot (Notes 1, 5, 7 & 8)	+25°C		25	40		25	40		25	40	%
Slew Rate (Notes 1, 5, 8 & 12)	+25°C	±4	±7		±4	±7		±4	±7		V/μs
Settling Time (Notes 1, 5, 8 & 12)	+25°C		1.5			1.5			1.5		μs
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	+25°C		3.0	3.7		3.0	4.0		3.0	4.0	mA
Power Supply Rejection Ratio (Note 9)	Full	80	90		74	90		74	90		dB

### TEST CONDITIONS

- NOTES: 1. R<sub>L</sub> = 2K  
 2. V<sub>CM</sub> = ±10V  
 3. V<sub>O</sub> < 90mV  
 4. V<sub>O</sub> = ±10V  
 5. C<sub>L</sub> = 100pF  
 6. V<sub>O</sub> = ±200mV

7. V<sub>O</sub> = ±200mV  
 8. See Transient response test circuits and waveforms  
 9. ΔV<sub>S</sub> = ±5V

10. This parameter value guaranteed by design calculations.  
 11. Full power bandwidth guaranteed by slew rate measurement.  
 FPBW = S.R./2πV<sub>peak</sub>  
 12. V<sub>OUT</sub> = ±5V

LINEAR

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# HARRIS

# HA-2620/2622/2625

## Very Wide Band, Uncompensated Operational Amplifiers

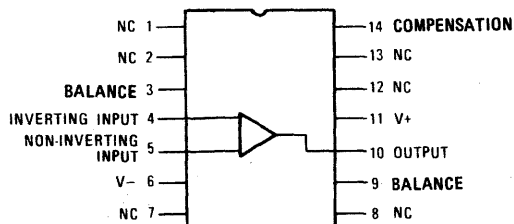
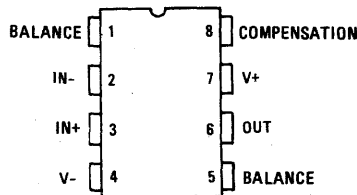
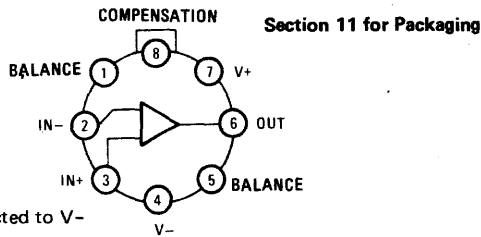
### FEATURES

- GAIN BANDWIDTH PRODUCT ( $A_V = 5$ ) 100MHz
- HIGH INPUT IMPEDANCE 500M $\Omega$
- LOW INPUT BIAS CURRENT 1nA
- LOW INPUT OFFSET CURRENT 1nA
- LOW INPUT OFFSET VOLTAGE 0.5mV
- HIGH GAIN 150K V/V
- HIGH SLEW RATE 35V/ $\mu$ s
- OUTPUT SHORT CIRCUIT PROTECTION

### APPLICATIONS

- VIDEO AND R.F. AMPLIFIERS
- PULSE AMPLIFIER
- AUDIO AMPLIFIERS AND FILTERS
- HIGH-Q ACTIVE FILTERS
- HIGH-SPEED COMPARATORS
- LOW DISTORTION OSCILLATORS

### PINOUT



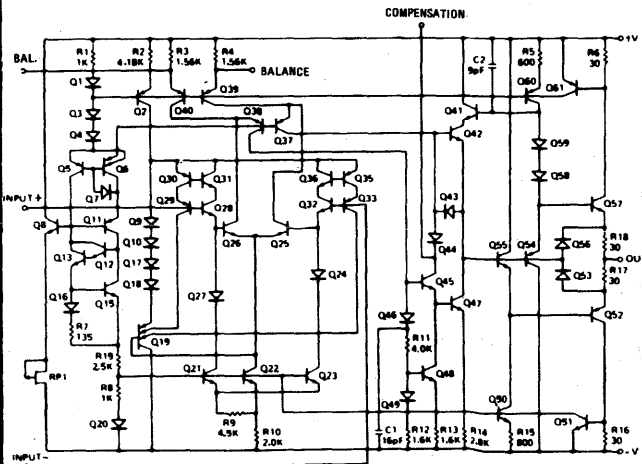
### DESCRIPTION

HA-2620/2622/2625 are bipolar operational amplifiers that feature very high input impedance (500M $\Omega$ , HA-2620) coupled with wideband AC performance. The high resistance of the input stage is complemented by low offset voltage (0.5mV, HA-2620) and low bias and offset current (1nA, HA-2620) to facilitate accurate signal processing. Input offset can be reduced further by means of an external nulling potentiometer. 100MHz gain-bandwidth product (HA-2620/2622/2625 are stable for closed loop gains greater than 5), 35V/ $\mu$ s slew rate and 150,000V/V open-loop gain enables HA-2620/2622/2625 to perform high-gain amplification of very fast, wideband signals. These dynamic characteristics, coupled with fast settling times, make these amplifiers ideally suited to pulse amplification designs as well as high frequency (e.g. video) applications. The frequency response of the amplifier can be tailored to exact design requirements by means of an external bandwidth control capacitor.

In addition to its application in pulse and video amplifier designs HA-2620/2622/2625 is particularly suited to other high performance designs such as high-gain low distortion audio amplifiers, high-Q and wideband active filters and high-speed comparators.

HA-2620 and HA-2622 are guaranteed over -55 $^{\circ}$ C to +125 $^{\circ}$ C. HA-2625 is specified from 0 $^{\circ}$ C to +75 $^{\circ}$ C. All devices are available in TO-99 cans, and 14 lead D.I.P. packages.

### SCHEMATIC



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# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V <sup>+</sup> and V <sup>-</sup> Terminals	45.0V
Differential Input Voltage	±12.0V
Peak Output Current	Full Short Circuit Protection
Internal Power Dissipation	300mW
Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C

## ELECTRICAL CHARACTERISTICS

V<sup>+</sup> = +15 VDC, V<sup>-</sup> = -15 VDC

PARAMETER	TEMPERATURE	HA-2620 -55°C to +125°C			HA-2622 -55°C to +125°C			HA-2625 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage (Note 1)	+25°C Full		0.5 4 6			3 5 7			3 5 7		mV mV
Bias Current	+25°C Full		1 10	15 35		5 25 60			5 25 40		nA nA
Offset Current	+25°C Full		1 5	15 35		5 25 60			5 25 40		nA nA
Input Resistance (Note 11)	+25°C	65	500		40	300		40	300		MΩ
Common Mode Range	Full	±11.0			±11.0			±11.0			V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Notes 2 & 3)	+25°C Full	100K 70K	150K		80K 60K	150K		80K 70K	150K		V/V V/V
Common Mode Rejection Ratio (Note 4)	Full	80	100		74	100		74	100		dB
Gain Bandwidth Product (Notes 2, 5, & 6)	+25°C		100			100			100		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 2)	Full	±10.0	±12.0		±10.0	±12.0		±10.0	±12.0		V
Output Current (Note 3)	+25°C	±15	±22		±10	±18		±10	±18		mA
Full Power Bandwidth (Notes 2, 3, 7 & 12)	+25°C	400	600		320	600		320	600		kHz
<b>TRANSIENT RESPONSE</b>											
Rise Time (Notes 2, 7 & 8)	+25°C		17	45		17	45		17	45	ns
Slew Rate (Notes 2, 7, 8 & 10)	+25°C	±25	±35		±20	±35		±20	±35		V/μs
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	+25°C		3.0	3.7		3.0	4.0		3.0	4.0	mA
Power Supply Rejection Ratio (Note 9)	Full	80	90		74	90		74	90		dB

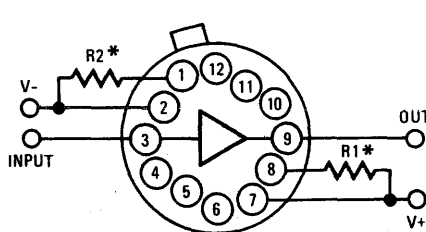
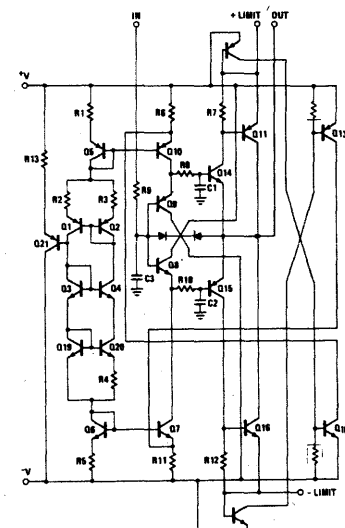
- NOTES: 1. Offset may be externally adjusted to zero.  
 2. R<sub>L</sub> = 2KΩ, C<sub>L</sub> = 50pF  
 3. V<sub>O</sub> = ±10.0V  
 4. V<sub>CM</sub> = ±10V  
 5. V<sub>O</sub> < 90mV  
 6. 40dB Gain  
 7. See transient response test circuits and waveforms

8. A<sub>V</sub> = 5 (The HA-2620 family is not stable at unity gain without external compensation.)  
 9. ΔV<sub>Sup</sub> = ±5V  
 10. V<sub>OUT</sub> = ±5V.  
 11. This parameter value based upon design calculations.  
 12. Full power bandwidth guaranteed based upon slew rate measurement  
 FPBW = S.R./2πV<sub>peak</sub>.

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FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>• OUTPUT CURRENT <math>\pm 400\text{mA}</math></li> <li>• SLEW RATE <math>500\text{V}/\mu\text{s}</math></li> <li>• BANDWIDTH 8MHz</li> <li>• FULL POWER BANDWIDTH 8MHz</li> <li>• INPUT RESISTANCE <math>2.0 \times 10^6 \Omega</math></li> <li>• OUTPUT RESISTANCE <math>2.0 \Omega</math></li> <li>• POWER SUPPLY RANGE <math>\pm 5\text{V}</math> to <math>\pm 20\text{V}</math></li> <li>• PACKAGE IS ELECTRICALLY ISOLATED</li> </ul>	<p>HA-2630 and HA-2635 are monolithic, unity voltage gain current amplifiers delivering extremely high slew rate, wide bandwidth, and full power bandwidth even under heavy output loading conditions. This dielectrically isolated current booster also offers high input impedance and low output resistance. These devices are intended to be used in series with an operational amplifier and inside the feedback loop whenever additional output current is required. Output current levels are programmable by selecting two optional external resistors.</p>
APPLICATIONS	<p>These current amplifiers offer an exceptional <math>500\text{V}/\mu\text{s}</math> slew rate and 8MHz bandwidth which allows them to be used with many high performance op amps in precision data recording and high speed coaxial cable driver designs. <math>2.0\text{M}</math> ohm input resistance and <math>2</math> ohm output resistance coupled with <math>\pm 400\text{mA}</math> output current make HA-2630 and HA-2635 ideal components in high fidelity audio output amplifier designs.</p> <p>HA-2630 and HA-2635 are available in an electrically isolated TO-8 type can for ease of mounting with or without a heat sink. HA-2630 is specified over the <math>-55^\circ\text{C}</math> to <math>+125^\circ\text{C}</math> range. HA-2635 is specified from <math>0^\circ\text{C}</math> to <math>+75^\circ\text{C}</math>.</p>
<ul style="list-style-type: none"> <li>• COAXIAL CABLE DRIVERS</li> <li>• AUDIO OUTPUT AMPLIFIERS</li> <li>• SERVO MOTOR DRIVERS</li> <li>• POWER SUPPLIES (BIPOLAR)</li> <li>• PRECISION DATA RECORDING</li> </ul>	
PINOUT	SCHEMATIC
<p style="text-align: center;">Section 11 for Packaging</p> <p style="text-align: center;">TOP VIEW</p>  <p style="text-align: center;">* Optional Current Limiting Resistor</p>	

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V- Terminals	40V	Operating Temperature Range:	
Input Voltage Range	± V Supply	-55°C ≤ T <sub>A</sub> ≤ +125°C	(HA-2630)
Output Current (Note 2)	±700mA	0°C ≤ T <sub>A</sub> ≤ +75°C	(HA-2635)
Internal Power Dissipation (Note 6)	Free Air: 1W	Storage Temperature Range:	
	In Heat Sink: 4W	-65°C ≤ T <sub>A</sub> ≤ +150°C	

## ELECTRICAL CHARACTERISTICS

V<sub>Supply</sub> = ±15 Volts      R<sub>L</sub> = 50 Ohms      R<sub>1</sub> = R<sub>2</sub> = 0 Ohms      Unless otherwise specified.

PARAMETER	TEMP.	HA-2630 -55°C to +125°C			HA-2635 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>								
Bias Current	+25°C Full		30	150 200		30	150 200	μA μA
Input Resistance	+25°C		2.0			2.0		MΩ
Input Capacitance	+25°C		5.0			5.0		pF
<b>TRANSFER CHARACTERISTICS</b>								
Voltage Gain (Note 1)	Full	.85	.95		.85	.95		V/V
Offset Voltage (V <sub>OUT</sub> - V <sub>IN</sub> )	+25°C Full		70	±200 ±300		70	±200 ±300	mV mV
Bandwidth (-3dB)	+25°C		8.0			8.0		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing	Full	±10	±12		±10	±12		V
Output Current (Note 1)	Full	±300	±400		±300	±400		mA
Output Resistance	+25°C		2.0			2.0		Ω
Full Power Bandwidth (Note 1)	+25°C		8.0			8.0		MHz
<b>TRANSIENT RESPONSE</b>								
Rise Time (Note 3)	+25°C		30			30		ns
Slew Rate (Note 4)	+25°C	200	500		200	500		V/μs
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	Full		15	20		15	23	mA
Supply Voltage Range	Full	±5		±20	±5		±20	V
Power Supply Rejection Ratio (Note 5)	Full		66			66		dB

- NOTES: 1. V<sub>O</sub> = ±10V  
 2. Heat sink is required for continuous short circuit protection, regardless of current limit setting.  
 3. V<sub>O</sub> = 0.4V p-p.  
 4. V<sub>O</sub> = 10V p-p.  
 5. ΔV<sub>SUPPLY</sub> = ±5V.  
 6. Without heat sink, derate by 14mW/°C ambient temperature above 100°C ambient, with heat sink, derate by 67mW/°C case temperature above 115°C case.

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# HARRIS

## HA-2640/2645

### High Voltage Operational Amplifier

#### FEATURES

- OUTPUT VOLTAGE SWING  $\pm 35V$
- SUPPLY VOLTAGE  $\pm 10V$  TO  $\pm 40V$
- OFFSET CURRENT  $5nA$
- BANDWIDTH  $4MHz$
- SLEW RATE  $5V/\mu s$
- COMMON MODE INPUT VOLTAGE SWING  $\pm 35V$
- OUTPUT OVERLOAD PROTECTION

#### DESCRIPTION

HA-2640 and HA-2645 are monolithic operational amplifiers which are designed to deliver unprecedented dynamic specifications for a high voltage internally compensated device. These dielectrically isolated devices offer very low values for offset voltage and offset current coupled with large output voltage swing and common mode input voltage.

For maximum reliability, these amplifiers offer unconditional output overload protection through current limiting and a chip temperature sensing circuit. This sensing device turns the amplifier "off", when the chip reaches a certain temperature level.

#### APPLICATIONS

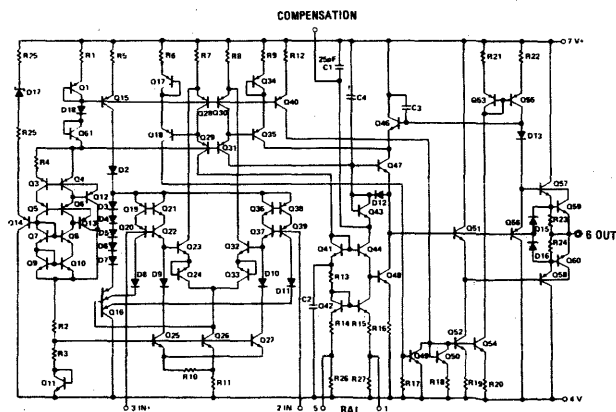
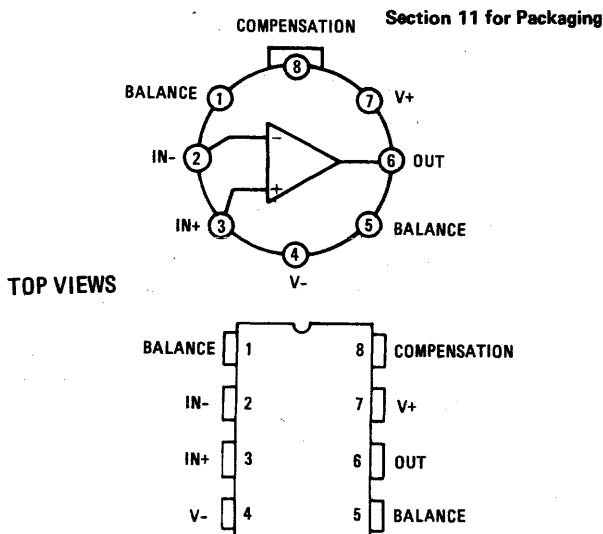
- INDUSTRIAL CONTROL SYSTEMS
- POWER SUPPLIES
- HIGH VOLTAGE REGULATORS
- RESOLVER EXCITATION
- SIGNAL CONDITIONING

These amplifiers deliver  $\pm 35V$  common mode input voltage swing,  $\pm 35V$  output voltage swing, and up to  $\pm 40V$  supply range for use in such designs as regulators, power supplies, and industrial control systems.  $4MHz$  gain bandwidth and  $5V/\mu s$  slew rate make these devices excellent components for high performance signal conditioning applications. Outstanding input and output voltage swings coupled with a low  $5nA$  offset current make these amplifiers excellent components for resolver excitation designs.

HA-2640 and HA-2645 are available in metal can (TO-99) packages and can be used as high performance pin-to-pin replacements for many general purpose op amps. HA-2640 is specified from  $-55^{\circ}C$  to  $+125^{\circ}C$  and HA-2645 is specified over the  $0^{\circ}C$  to  $+75^{\circ}C$  range.

#### PINOUT

#### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V- Terminals 100V  
 Input Voltage Range  $\pm 37V$   
 Output Current/Full Short Circuit Protection  
 Internal Power Dissipation 680mW\*

Operating Temperature Range  
 $-55^{\circ}C \leq T_A \leq +125^{\circ}C$  (HA-2640)  
 $0^{\circ}C \leq T_A \leq +75^{\circ}C$  (HA-2645)  
 Storage Temperature Range  
 $-65^{\circ}C \leq T_A \leq +150^{\circ}C$

\*Derate by 4.6mW/ $^{\circ}C$  above +25 $^{\circ}C$

## ELECTRICAL CHARACTERISTICS

V<sub>Supply</sub> =  $\pm 40V$ , R<sub>L</sub> = 5K, Unless Otherwise Specified.

PARAMETER	TEMP.	HA-2640 -55 $^{\circ}C$ to +125 $^{\circ}C$			HA-2645 0 $^{\circ}C$ to +75 $^{\circ}C$			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25 $^{\circ}C$		2	4		2	6	mV
	Full			6			7	mV
Offset Voltage Average Drift	Full		15			15		$\mu V/^{\circ}C$
Bias Current	+25 $^{\circ}C$		10	25		12	30	nA
	Full			50			50	nA
Offset Current	+25 $^{\circ}C$		5	12		15	30	nA
	Full			35			50	nA
Input Resistance (Note 10)	+25 $^{\circ}C$	50	250		40	200		M $\Omega$
Common Mode Range	Full	$\pm 35$			$\pm 35$			V
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 8)	+25 $^{\circ}C$	100K	200K		100K	200K		V/V
	Full	75K			75K			V/V
Common Mode Rejection Ratio (Note 1)	Full	80	100		74	100		dB
Unity Gain Bandwidth (Note 2)	+25 $^{\circ}C$		4			4		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing	Full	$\pm 35$			$\pm 35$			V
Output Current (Note 9)	+25 $^{\circ}C$	$\pm 12$	$\pm 15$		$\pm 10$	$\pm 12$		mA
Output Resistance	+25 $^{\circ}C$		500			500		$\Omega$
Full Power Bandwidth (Notes 3 & 11)	+25 $^{\circ}C$		23			23		kHz
<b>TRANSIENT RESPONSE (Note 7)</b>								
Rise Time (Notes 4, 6)	+25 $^{\circ}C$		60	100		60	100	ns
Overshoot (Notes 4, 6)	+25 $^{\circ}C$		15	30		15	40	%
Slew Rate (Note 6)	+25 $^{\circ}C$	$\pm 3$	$\pm 5$		$\pm 2.5$	$\pm 5$		V/ $\mu s$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	+25 $^{\circ}C$		3.2	3.8		3.2	4.5	mA
Supply Voltage Range	Full	$\pm 10$		$\pm 40$	$\pm 10$		$\pm 40$	V
Power Supply Rejection Ratio (Note 5)	Full	80	90		74	90		dB

- NOTES: 1. V<sub>CM</sub> =  $\pm 20V$   
 2. V<sub>O</sub> = 90mV  
 3. V<sub>O</sub> =  $\pm 35V$   
 4. V<sub>O</sub> =  $\pm 200mV$

5. V<sub>S</sub> =  $\pm 10V$  to  $\pm 40V$   
 6. A<sub>V</sub> = 1  
 7. C<sub>L</sub> = 50pF  
 8. V<sub>O</sub> =  $\pm 30V$   
 9. R<sub>L</sub> = 1K

10. This parameter based upon design calculations.  
 11. Full power bandwidth guaranteed based upon slew rate measurement.  
 FPBW = S.R./2 $\pi$ V<sub>peak</sub>.



# HA-2650/2655

## Dual High Performance Operational Amplifier

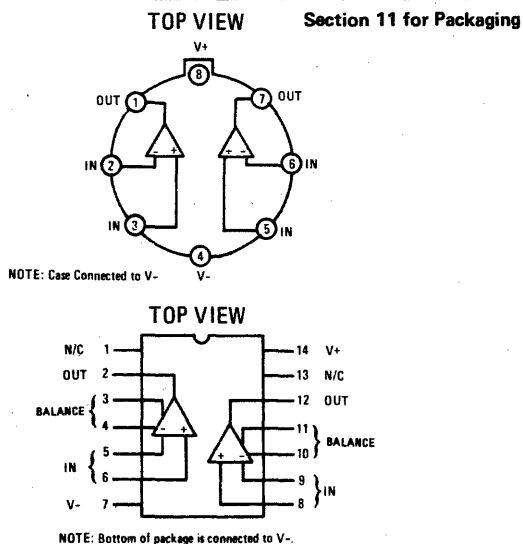
### FEATURES

- SLEW RATE 5V/μs
- BANDWIDTH 8MHz
- BIAS CURRENT 35nA
- AV. OFFSET VOLTAGE DRIFT 8μV/°C
- POWER CONSUMPTION 75mW
- SUPPLY VOLTAGE RANGE ±2V TO ±20V

### APPLICATIONS

- VIDEO AMPLIFIERS
- HIGH IMPEDANCE, WIDEBAND BUFFERS
- INTEGRATORS
- AUDIO AMPLIFIERS
- ACTIVE FILTERS

### PINOUT



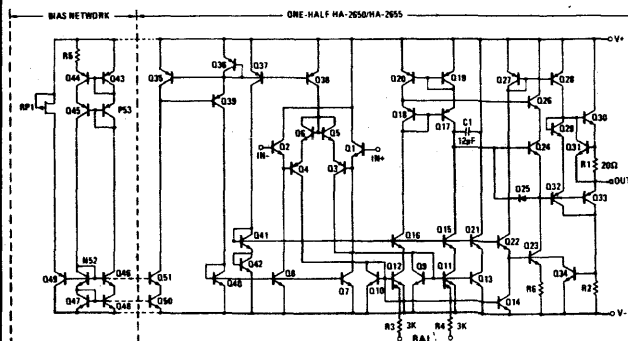
### DESCRIPTION

HA-2650/2655 contains two internally compensated operational amplifiers offering high slew rate and high frequency performance combined with exceptional DC characteristics. 5V/μ sec slew rate and 8MHz bandwidth make these amplifiers suitable for processing fast, wideband signals extending into the video frequency spectrum. Signal processing accuracy is enhanced by front-end performance that includes 1.5mV offset voltage, 8μ V/°C offset voltage drift and low offset and bias current (1nA and 35nA respectively). Offset voltage can be trimmed to zero on the devices offered in dual-in-line packages. Signal conditioning is further enhanced by 500MΩ input impedance.

Applications for HA-2650/2655 include video circuit designs such as high impedance buffers, integrators, tone generators and filters. These amplifiers are also ideal components for active filtering of audio and voice signals.

HA-2650/2655 are offered in 14 pin D.I.P. and metal TO-99 packages and are also available in dice form. HA-2650 is specified from -55°C to +125°C. HA-2655 operates from 0°C to +75°C.

### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

$T_A = +25^{\circ}\text{C}$  Unless Otherwise Stated

Power Dissipation (Note 2) TO-99 300 mW  
TO-116 300 mW

Voltage Between V+ and V- Terminals 40.0V  
Differential Input Voltage  $\pm 30.0\text{V}$   
Input Voltage (Note 1)  $\pm 15.0\text{V}$   
Output Short Circuit Duration Indefinite

Operating Temperature Range:  
HA-2650  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$   
HA-2655  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$   
Storage Temperature Range  $-65^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$

## ELECTRICAL CHARACTERISTICS

V+ = 15V V- = -15V

PARAMETER	TEMP.	HA-2650 -55°C to +125°C			HA-2655 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		1.5	3		2	5	mV
	Full			5			7	mV
Av. Offset Voltage Drift	Full		8			8		$\mu\text{V}/^{\circ}\text{C}$
Bias Current	+25°C		35	100		50	200	nA
	Full			200			300	nA
Offset Current	+25°C		1	30		2	60	nA
	Full			60			100	nA
Common Mode Range	Full	$\pm 13$			$\pm 13$			V
Differential Input Resistance (Note 9)	+25°C	5	20		5	20		M $\Omega$
Common Mode Input Resistance	+25°C		500			500		M $\Omega$
Input Capacitance	+25°C		5			5		pF
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 3ab)	+25°C	20K	40K		15K	40K		V/V
	Full	15K			10K			V/V
Common Mode Rejection Ratio (Note 4)	+25°C	80	100		74	100		dB
	Full	80			74			dB
<b>OUTPUT CHARACTERISTICS</b>								
*Output Voltage Swing (Note 3c)	+25°C	$\pm 13$	$\pm 14$		$\pm 13$	$\pm 14$		V
	Full	$\pm 13$			$\pm 13$			V
Full Power Bandwidth (Notes 5 & 10)	+25°C	30	80		30	80		KHz
Output Current (Note 3a)	+25°C		$\pm 20$			$\pm 18$		mA
Output Resistance	+25°C		100			100		$\Omega$
<b>TRANSIENT RESPONSE (Note 6)</b>								
Rise Time (Note 7)	+25°C		40	80		40	90	ns
Overshoot (Note 7)	+25°C		15	30		15	40	%
Slew Rate	+25°C	$\pm 2$	$\pm 5$		$\pm 2$	$\pm 5$		V/ $\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	+25°C		2.5	4		3	5	mA
Power Supply Rejection Ratio (Note 8)	+25°C	80	100		74	100		dB
	Full	80			74			dB

NOTES: 1. For supply voltages less than  $\pm 15\text{V}$ , the absolute maximum input voltage is equal to the supply voltage.  
2. Derate at  $4.7\text{mW}/^{\circ}\text{C}$  at ambient temperatures above  $+110^{\circ}\text{C}$ .  
3. (a)  $V_O = \pm 10\text{V}$  (b)  $R_L = 2\text{K}$  (c)  $R_L = 10\text{K}$

4.  $V_{CM} = \pm 5.0\text{V}$   
5.  $A_V = 1, R_L = 2\text{K}, V_O = 20\text{V}_{pp}$   
6. See transient response/slew rate circuit.  
7.  $V_{in} = 200\text{mV}$   
8.  $\Delta V = \pm 5.0\text{V}$

9. This parameter value based upon design calculations.  
10. Full power bandwidth guaranteed based upon slew rate measurement  $\text{FPBW} = \text{S.R.}/2\pi V_{peak}$ .

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# HA-2720/25

## Wide Range Programmable Operational Amplifier

### FEATURES

- WIDE PROGRAMMING RANGE
 

SLEW RATE	0.06 TO 6V/ $\mu$ s
BANDWIDTH	5kHz TO 10MHz
BIAS CURRENT	0.4 TO 50nA
SUPPLY CURRENT	1 $\mu$ A TO 1.5mA
- WIDE POWER SUPPLY RANGE  $\pm 1.2$  TO  $\pm 18$ V
- CONSTANT AC PERFORMANCE OVER SUPPLY RANGE

### APPLICATIONS

- ACTIVE FILTERS
- CURRENT CONTROLLED OSCILLATORS
- VARIABLE ACTIVE FILTERS
- MODULATORS
- BATTERY-POWERED EQUIPMENT

### DESCRIPTION

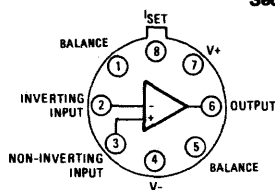
HA-2720/2725 programmable amplifiers are internally compensated monolithic devices offering a wide range of performance, that can be controlled by adjusting the circuits' "set" current (I<sub>SET</sub>). By means of adjusting an external resistor or current source, power dissipation, slew rate, bandwidth, output current and input noise can be programmed to desired levels. This versatile adjustment capability enables HA-2720/2725 to provide optimum design solutions by delivering the required level of performance with minimum possible power dissipation. HA-2720 and HA-2725 can, therefore, be utilized as the standard amplifier for a variety of designs simply by adjusting their programming current.

A major advantage of HA-2720/2725 is that operating characteristics remain virtually constant over a wide supply range ( $\pm 1.2$ V to  $\pm 15$ V), allowing the amplifiers to offer maximum performance in almost any system including battery-operated equipment. A primary application for HA-2720/2725 is in active filters for a wide variety of signals that differ in frequency and amplitude. Also, by modulating the "set" current, HA-2720/2725 can be used for designs such as current controlled oscillators modulators, sample and hold circuits and variable active filters.

HA-2720 is guaranteed over  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ . HA-2725 is specified from  $0^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$ . Both parts are available in TO-99 cans or dice form.

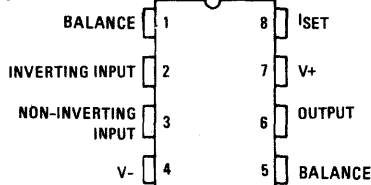
### PINOUT

Section 11 for Packaging

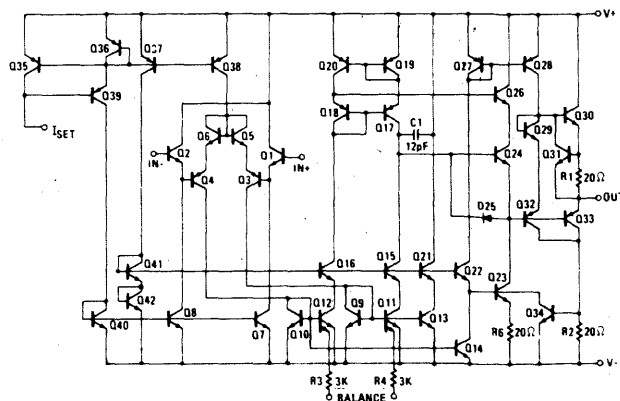


NOTE: Case tied to V<sup>-</sup>

TOP VIEWS



### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V- Terminals	45.0V	Power Dissipation (Note 2)	300mW
Differential Input Voltage	±30.0V	Operating Temperature Range:	
Input Voltage (Note 1)	±15.0V	HA-2720	-55°C ≤ T <sub>A</sub> ≤ +125°C
I <sub>SET</sub> (Current at I <sub>SET</sub> )	500μA	HA-2725	0°C ≤ T <sub>A</sub> ≤ +75°C
V <sub>SET</sub> (Voltage to Gnd. at I <sub>SET</sub> )	V+ - 2.0V ≤ V <sub>SET</sub> ≤ V+	Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C

## ELECTRICAL CHARACTERISTICS

V+ = +3.0V, V- = -3.0V

PARAMETER	TEMP.	HA-2720 -55°C to +125°C						HA-2725 0°C to +75°C						UNITS
		I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	25°C		2.0	3.0		2.0	3.0		2.0	5.0		2.0	5.0	mV
	Full			5.0		5.0		7.0		7.0		7.0		mV
Offset Current	25°C		0.5	3.0		1.0	10		0.5	5.0		1.0	10	nA
	Full			7.5		20		7.5		20		20		nA
Bias Current	25°C		2.0	5.0		8.0	20		2.0	10		8.0	30	nA
	Full			10		40		10		40		40		nA
Input Resistance (Note 10)	25°C		50			5			50			5		MΩ
Input Capacitance	25°C		3.0			3.0			3.0			3.0		pF
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 9)	25°C	15K	40K		15K	40K		15K	40K		15K	40K		V/V
	Full	10K			10K			10K			10K			V/V
Common Mode Rejection Ratio (Note 4)	Full	80			80			74			74			dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 3)	25°C	±2.0	±2.2		±2.0	±2.2		±2.0	±2.2		±2.0	±2.2		V
	Full	±2.0			±1.9			±2.0			±2.0			V
Output Current (Note 5)	25°C		±0.2			±2.0			±0.2			±2.0		mA
Output Resistance	25°C		2K			500			2K			500		Ω
Output Short-Circuit Current	25°C		2.8			14			2.8			14		mA
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 6)	25°C		2.5			0.25			2.5			0.25		μs
Overshoot (Note 6)	25°C		5			10			5			10		%
Slew Rate (Note 7)	25°C		0.07			0.70			0.07			0.70		V/μs
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current	25°C		15			170			15			170		μA
	Full			25		250			25			250		μA
Power Supply Rejection Ratio (Note 8)	Full	100			100			150			150			μV/V

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**ELECTRICAL CHARACTERISTICS**

V+ = +15.0V, V- = -15.0V

PARAMETER	TEMP.	HA-2720 -55°C to +125°C						HA-2725 0°C to +75°C						UNITS
		I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	25°C Full		2.0 3.0	3.0 5.0		2.0 3.0	3.0 5.0		2.0 5.0	5.0 7.0		2.0 5.0	5.0 7.0	mV mV
Offset Current	25°C Full		0.5 3.0	3.0 7.5		1.0 10	10 20		0.5 5.0	5.0 7.5		1.0 10	10 20	nA nA
Bias Current	25°C Full		2.0 5.0	5.0 10		8.0 20	20 40		2.0 10	10 10		8.0 30	30 40	nA nA
Input Resistance (Note 10)	25°C		50			5			50			5		MΩ
Input Capacitance	25°C		3.0			3.0			3.0			3.0		pF
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Notes 3 & 9)	25°C Full	30K 20K	100K		30K 20K	120K		25K 20K	100K		25K 20K	120K		V/V V/V
Common Mode Rejection Ratio (Note 4)	25°C Full	80	90		80	90		74	90		74	90		dB dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 3)	25°C Full	±12 ±10	±13.5		±12 ±10	±13.5		±12 ±10	±13.5		±12 ±10	±13.5		V V
Output Current (Note 5)	25°C		±0.5			±5.0			±0.5			±5.0		mA
Output Resistance	25°C		2K			500			2K			500		Ω
Output Short-Circuit Current	25°C		3.7			19			3.7			19		mA
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 6)	25°C		2.0			0.2			2.0			0.2		μs
Overshoot (Note 6)	25°C		5			15			5			15		%
Slew Rate (Note 7)	25°C		0.1			0.8			0.1			0.8		V/μs
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current	25°C Full		20 50			210 450			20 50			210 450		μA μA
Power Supply Rejection Ratio (Note 8)	Full	100			100			150			150			μV/V

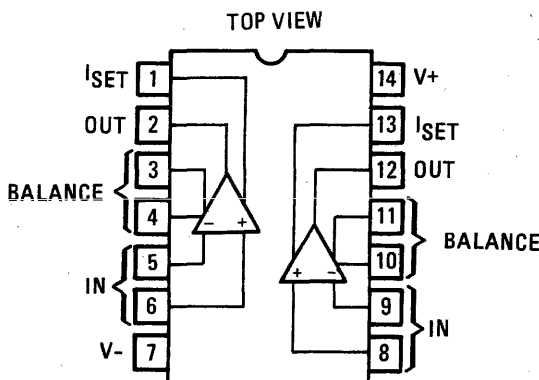
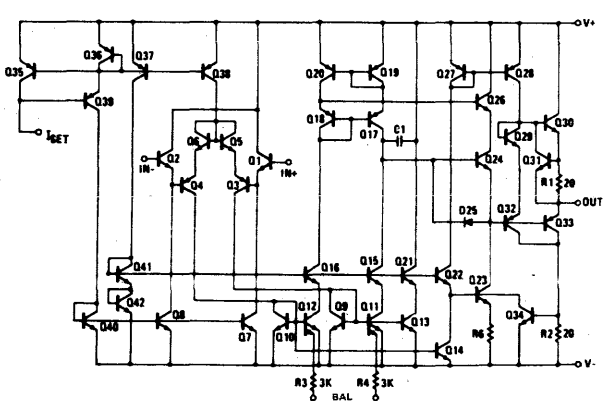
- NOTES: 1. For supply voltages less than ±15.0V, the absolute maximum input voltage is equal to supply voltage.  
 2. Derate at 6.8mW/°C for operation ambient temperatures above 75°C.

$V_{SUPPLY} = \pm 3.0V$	$V_{SUPPLY} = \pm 15.0V$	$I_{SET} = 1.5\mu A$	$I_{SET} = 15\mu A$
3. T = +25°C and Full	T = +25°C T = Full	R <sub>L</sub> = 75KΩ R <sub>L</sub> = 75KΩ	R <sub>L</sub> = 5KΩ R <sub>L</sub> = 75KΩ
4. V <sub>CM</sub> = ±1.5V	V <sub>CM</sub> = ±5.0V		
5. V <sub>O</sub> = ±2.0V	V <sub>O</sub> = ±10.0V		
6. ← A <sub>V</sub> = +1, V <sub>IN</sub> = 400mV, R <sub>L</sub> = 5K, C <sub>L</sub> = 100pF →			
7. V <sub>O</sub> = ±2.0V	V <sub>O</sub> = ±10.0V	R <sub>L</sub> = 20K	R <sub>L</sub> = 5K
8. ΔV = ±1.5V	ΔV = ±5.0V		
9. V <sub>O</sub> = ±1.0V	V <sub>O</sub> = ±10.0V		

10. This parameter based upon design calculations.

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## Wide Range Dual Programmable Operational Amplifier

FEATURES	DESCRIPTION										
<ul style="list-style-type: none"> <li>WIDE PROGRAMMING RANGE               <table border="0" data-bbox="349 577 836 724"> <tr> <td>SET CURRENT</td> <td>0.1 TO 100<math>\mu</math>A</td> </tr> <tr> <td>SLEW RATE</td> <td>0.06 TO 6V/<math>\mu</math>s</td> </tr> <tr> <td>BANDWIDTH</td> <td>5kHz TO 10MHz</td> </tr> <tr> <td>BIAS CURRENT</td> <td>0.4 TO 50nA</td> </tr> <tr> <td>SUPPLY CURRENT</td> <td>1<math>\mu</math>A TO 1.5mA</td> </tr> </table> </li> <li>WIDE POWER SUPPLY RANGE <math>\pm 1.2</math> TO <math>\pm 18</math>V</li> <li>CONSTANT AC PERFORMANCE OVER SUPPLY RANGE</li> </ul>	SET CURRENT	0.1 TO 100 $\mu$ A	SLEW RATE	0.06 TO 6V/ $\mu$ s	BANDWIDTH	5kHz TO 10MHz	BIAS CURRENT	0.4 TO 50nA	SUPPLY CURRENT	1 $\mu$ A TO 1.5mA	<p>HA-2730/2735 Dual Programmable Amplifiers are internally compensated monolithic devices offering a wide range of performance, that can be controlled by adjusting the circuits' "set" current (<math>I_{SET}</math>). By means of adjusting an external resistor or current source, power dissipation, slew rate, bandwidth, output current and input noise can be programmed to desired levels. Each amplifier on the chip can be adjusted independently. This versatile adjustment capability enables HA-2730/2735 to provide optimum design solutions by delivering the required level of performance with minimum possible power dissipation. HA-2730/2735 can, therefore, be utilized as the standard amplifier for a variety of designs simply by adjusting their programming current.</p> <p>A major advantage of HA-2730/2735 is that operating characteristics remain virtually constant over a wide supply range (<math>\pm 1.2</math>V to <math>\pm 15</math>V), allowing the amplifiers to offer maximum performance in almost any system including battery-operated equipment. A primary application for HA-2730/2735 is in active filters for a wide variety of signals that differ in frequency and amplitude. Also, by modulating the "set" current, HA-2730/2735 can be used for designs such as current controlled oscillators, modulators, sample and hold circuits and variable active filters.</p> <p>HA-2730 is guaranteed over <math>-55^{\circ}</math>C to <math>+125^{\circ}</math>C. HA-2735 is specified from <math>0^{\circ}</math>C to <math>+75^{\circ}</math>C. Both parts are available in 14 lead D.I.P. package or dice form:</p>
SET CURRENT	0.1 TO 100 $\mu$ A										
SLEW RATE	0.06 TO 6V/ $\mu$ s										
BANDWIDTH	5kHz TO 10MHz										
BIAS CURRENT	0.4 TO 50nA										
SUPPLY CURRENT	1 $\mu$ A TO 1.5mA										
<h3>APPLICATIONS</h3> <ul style="list-style-type: none"> <li>ACTIVE FILTERS</li> <li>CURRENT CONTROLLED OSCILLATORS</li> <li>VARIABLE ACTIVE FILTERS</li> <li>MODULATORS</li> <li>BATTERY-POWERED EQUIPMENT</li> </ul>											
<h3>PINOUT</h3> <p style="text-align: center;">Section 11 for Packaging</p> <p style="text-align: center;">TOP VIEW</p>  <p>NOTE: Bottom of package is connected to V-.</p>	<h3>SCHEMATIC</h3>  <p style="text-align: center;">(ONE HALF ONLY) HA-2730/35</p>										

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**ABSOLUTE MAXIMUM RATINGS**

Voltage Between V+ and V- Terminals	45.0V	Power Dissipation (Note 2)	500mW
Differential Input Voltage	±30.0V	Operating Temperature Range:	
Input Voltage (Note 1)	±15.0V	HA-2730	-55°C ≤ T <sub>A</sub> ≤ +125°C
I <sub>SET</sub> (Current at I <sub>SET</sub> )	500μA	HA-2735	0°C ≤ T <sub>A</sub> ≤ +75°C
V <sub>SET</sub> (Voltage to Gnd. at I <sub>SET</sub> )	V+ - 2.0V ≤ V <sub>SET</sub> ≤ V+	Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C

**ELECTRICAL CHARACTERISTICS (Each Side)**

V+ = +3.0V, V- = -3.0V

PARAMETER	TEMP.	HA-2730 -55°C to +125°C						HA-2735 0°C to +75°C						UNITS
		I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	25°C		2.0	3.0		2.0	3.0		2.0	5.0		2.0	5.0	mV
	Full			5.0		5.0		7.0		7.0		7.0	7.0	mV
Offset Current	25°C		0.5	3.0		1.0	10		0.5	5.0		1.0	10	nA
	Full			7.5		20		7.5		20		20	20	nA
Bias Current	25°C		2.0	5.0		8.0	20		2.0	10		8.0	30	nA
	Full			10		40		10		10		40	40	nA
Input Resistance (Note 10)	25°C		50			5			50			5		MΩ
Input Capacitance	25°C		3.0			3.0			3.0			3.0		pF
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Notes 3 & 9)	25°C	15K	40K		15K	40K		15K	40K		15K	40K		V/V
	Full	10K			10K			10K			10K			V/V
Common Mode Rejection Ratio (Note 4)	Full	80			80			74			74			dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 3)	25°C	±2.0	±2.2		±2.0	±2.2		±2.0	±2.2		±2.0	±2.2		V
	Full	±2.0			±1.9			±2.0			±2.0			V
Output Current (Note 5)	25°C		±0.2			±2.0			±0.2			±2.0		mA
Output Resistance	25°C		2K			500			2K			500		Ω
Output Short-Circuit Current	25°C		2.8			14			2.8			14		mA
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 6)	25°C		2.5			0.25			2.5			0.25		μs
Overshoot (Note 6)	25°C		5			10			5			10		%
Slew Rate (Note 7)	25°C		0.07			0.70			0.07			0.70		V/μs
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current (Each Amp)	25°C		15			170			15			170		μA
	Full			25		250			25			250		μA
Power Supply Rejection Ratio (Note 8)	Full	100			100			150			150			μV/V

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# SPECIFICATIONS



## ELECTRICAL CHARACTERISTICS (Each Side)

V+ = +15.0V, V- = -15.0V

PARAMETER	TEMP.	HA-2730 -55°C to +125°C						HA-2735 0°C to +75°C						UNITS
		I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			I <sub>SET</sub> = 1.5μA			I <sub>SET</sub> = 15μA			
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	25°C		2.0	3.0		2.0	3.0		2.0	5.0		2.0	5.0	mV
	Full			5.0		5.0		7.0		7.0		7.0		mV
Offset Current	25°C		0.5	3.0		1.0	10		0.5	5.0		1.0	10	nA
	Full			7.5		20		7.5		20		20		nA
Bias Current	25°C		2.0	5.0		8.0	20		2.0	10		8.0	30	nA
	Full			10		40		10		40		40		nA
Input Resistance (Note 10)	25°C		50			5			50			5		MΩ
Input Capacitance	25°C		3.0			3.0			3.0			3.0		pF
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Notes 3 & 9)	25°C	30K	100K		30K	120K		25K	100K		25K	120K		V/V
	Full	20K			20K			20K			20K			V/V
Common Mode Rejection Ratio (Note 4)	25°C		90			90			90			90		dB
	Full	80			80			74			74			dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 3)	25°C	±12	±13.5		±12	±13.5		±12	±13.5		±12	±13.5		V
	Full	±10			±10			±10			±10			V
Output Current (Note 5)	25°C		±0.5			±5.0			±0.5			±5.0		mA
Output Resistance	25°C		2K			500			2K			500		Ω
Output Short-Circuit Current	25°C		3.7			19			3.7			19		mA
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 6)	25°C		2.0			0.2			2.0			0.2		μs
Overshoot (Note 6)	25°C		5			15			5			15		%
Slew Rate (Note 7)	25°C		0.1			0.8			0.1			0.8		V/μs
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current (Each Amp)	25°C		20			210			20			210		μA
	Full			50		450			50			450		μA
Power Supply Rejection Ratio (Note 8)	Full	100			100			150			150			μV/V

NOTES: 1. For supply voltages less than ±15.0V, the absolute maximum input voltage is equal to supply voltage.  
2. Derate at 4.7mW/°C at ambient temperature above 68°C.

V <sub>SUPPLY</sub> = ±3.0V	V <sub>SUPPLY</sub> = ±15.0V	I <sub>SET</sub> = 1.5μA	I <sub>SET</sub> = 15μA
3. T = +25°C and Full	T = +25°C	R <sub>L</sub> = 75KΩ	R <sub>L</sub> = 5KΩ
	T = Full	R <sub>L</sub> = 75KΩ	R <sub>L</sub> = 75KΩ
4. V <sub>CM</sub> = ±1.5V	V <sub>CM</sub> = ±5.0V		
5. V <sub>O</sub> = ±2.0V	V <sub>O</sub> = ±10.0V		
6. $\leftarrow$ A <sub>V</sub> = +1, V <sub>IN</sub> = 400mV, R <sub>L</sub> = 5K, C <sub>L</sub> = 100pF $\rightarrow$		R <sub>L</sub> = 20K	R <sub>L</sub> = 5K
7. V <sub>O</sub> = ±2.0V	V <sub>O</sub> = ±10.0V		
8. ΔV = ±1.5V	ΔV = ±5.0V		
9. V <sub>O</sub> = ±1.0V	V <sub>O</sub> = ±10.0V		

10. This parameter value based upon design calculations.

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# HA-2740

## Quad Programmable Operational Amplifier

### FEATURES

- WIDE PROGRAMMING RANGE
  - ▶ SLEW RATE 0.8V/μs
  - ▶ BANDWIDTH 1MHz
  - ▶ BIAS CURRENT 8nA
  - ▶ SUPPLY CURRENT 250μA
- WIDE POWER SUPPLY RANGE
- CONSTANT AC PERFORMANCE OVER SUPPLY RANGE

### APPLICATIONS

- ACTIVE FILTERS
- CURRENT CONTROLLED OSCILLATORS
- VARIABLE ACTIVE FILTERS
- MODULATORS
- BATTERY-POWERED EQUIPMENT

### DESCRIPTION

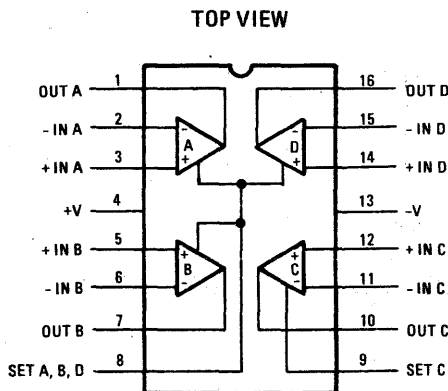
The Harris HA-2740 programmable amplifier is an internally compensated monolithic device offering a wide range of performance, that can be controlled by adjusting the circuit "set" current (ISET). By means of adjusting an external resistor or current source, power dissipation, slew rate, bandwidth, output current and input noise can be programmed to desired levels. This versatile adjustment capability enables the HA-2740 to provide optimum design solutions by delivering the required level of performance with minimum possible power dissipation. The HA-2740 can, therefore, be utilized as the standard amplifier for a variety of designs simply by adjusting programming current.

A major advantage of the HA-2740 is that operating characteristics remain virtually constant over a wide supply range ( $\pm 1.2V$  to  $\pm 18V$ ), allowing the amplifier to offer maximum performance in almost any system including battery-operated equipment. A primary application for the HA-2740 is in active filters for a wide variety of signals that differ in frequency and amplitude. Also, by modulating the "set" current, the HA-2740 can be used for designs such as current controlled oscillators, modulators, sample and hold circuits and variable active filters.

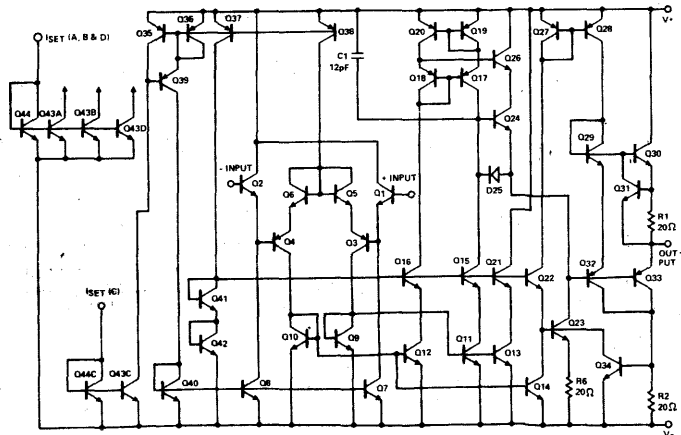
The HA-2740-2 is guaranteed over  $-55^{\circ}C$  to  $+125^{\circ}C$ . The HA-2740-5 is specified from  $0^{\circ}C$  to  $+75^{\circ}C$ . Both parts are available in a 16 pin dual-in-line package.

### PINOUT

Section 11 for Packaging



### SCHEMATIC



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**ABSOLUTE MAXIMUM RATINGS** (Note 1)

Voltage Between V+ and V- Terminals	45.0V	Power Dissipation	300mW
Differential Input Voltage	± 30.0V	Operating Temperature Range:	
Input Voltage (Note 2)	± 15.0V	HA-2740-2	-55°C ≤ T <sub>A</sub> ≤ +125°C
I <sub>SET</sub> (Current at I <sub>SET</sub> )	500µA	HA-2740-5	0°C ≤ T <sub>A</sub> ≤ +75°C
V <sub>SET</sub> (Voltage to Gnd. at I <sub>SET</sub> )	V+ - 2.0 ≤ V <sub>SET</sub> ≤ V+	Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C

**ELECTRICAL CHARACTERISTICS** V+ = +15.0V, V- = -15.0V

PARAMETER	TEMP	HA-2740-2 -55°C to +125°C						HA-2740-5 0°C to +75°C						UNITS
		I <sub>SET</sub> = 1.5µA			I <sub>SET</sub> = 15µA			I <sub>SET</sub> = 1.5µA			I <sub>SET</sub> = 15µA			
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	25°C		2.0	3.0		2.0	3.0		2.0	5.0		2.0	5.0	mV
	Full			5.0			5.0			7.0			7.0	mV
Offset Current	25°C		0.5	3.0		1.0	10		0.5	5.0		1.0	10	nA
	Full			7.5			20			7.5			30	nA
Bias Current	25°C		2.0	5.0		8.0	20		2.0	10		8.0	30	nA
	Full			10			40			10			40	nA
Input Resistance (Note 3)	25°C		50			5			50			5		MΩ
Input Capacitance	25°C		3.0			3.0			3.0			3.0		pF
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 4)	25°C	30K	100K		30K	120K		25K	100K		25K	120K		V/V
	Full	20K			20K			20K			20K			V/V
Common Mode Rejection Ratio (Note 5)	25°C		100			100			100			100		dB
	Full	80			80			74			74			dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 6)	25°C	± 12	± 14		± 12	± 14		± 12	± 14		± 12	± 14		V
	Full	± 10			± 10			± 10			± 10			V
Channel Separation (Note 7)	25°C		110			110			110			110		dB
Output Current (Note 8)	25°C		± 0.5			± 5.0			± 0.5			± 5.0		mA
Output Resistance	25°C		2K			500			2K			500		Ω
Output Short Circuit Current	25°C		3.6			16			3.6			16		mA
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 9)	25°C		2.0			0.2			2.0			0.2		µs
Overshoot (Note 9)	25°C		2			10			2			10		%
Slew Rate (Note 10)	25°C		0.1			0.8			0.1			0.8		V/µs
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current (each amp)	25°C		25			250			25			250		µA
	Full			50			450			50			450	µA
Power Supply Rejection Ratio (Note 11)	Full	100				100		150			150			µV/V

**NOTES:**

- Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- For supply voltages less than ± 15V, the absolute maximum input voltage is equal to the supply voltage.
- This parameter based upon design calculations.
- V<sub>O</sub> = ± 10V, R<sub>L</sub> = 5K @ I<sub>SET</sub> = 15µA  
R<sub>L</sub> = 75K @ I<sub>SET</sub> = 1.5µA
- V<sub>CM</sub> = ± 5V
- R<sub>L</sub> = 5kΩ @ I<sub>SET</sub> = 15µA, R<sub>L</sub> = 75K @ I<sub>SET</sub> = 1.5µA
- R<sub>S</sub> = 1kΩ, f = 100Hz.
- V<sub>O</sub> = ± 10V
- A<sub>V</sub> = 1, V<sub>IN</sub> = 200r.v., R<sub>L</sub> = 5k, C<sub>L</sub> = 100pF.
- V<sub>O</sub> = ± 10V, R<sub>L</sub> = 5K @ I<sub>SET</sub> = 15µA,  
R<sub>L</sub> = 20K @ I<sub>SET</sub> = 1.5µA
- A<sub>V</sub> = ± 5V.

## High Performance Quad Operational Amplifier

### FEATURES

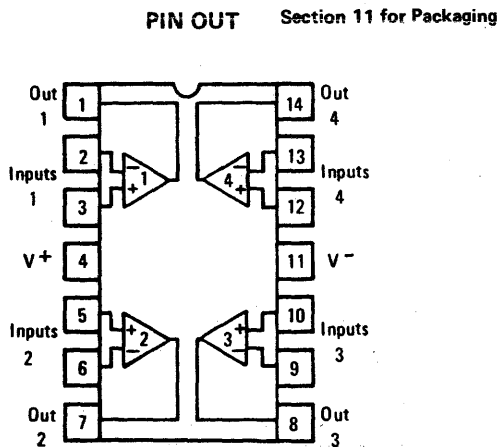
- SLEW RATE 1.6 V/ $\mu$ S (TYP.)
- BANDWIDTH 3.5 MHz (TYP.)
- INPUT VOLTAGE NOISE ( $f = 1\text{kHz}$ ) 9 NV/ $\sqrt{\text{Hz}}$  (TYP.)
- INPUT OFFSET VOLTAGE 0.5 mV (TYP.)
- INPUT BIAS CURRENT 60 nA (TYP.)
- SUPPLY RANGE  $\pm 2\text{V}$  to  $\pm 20\text{V}$
- NO CROSSOVER DISTORTION
- STANDARD QUAD PIN-OUT

### DESCRIPTION

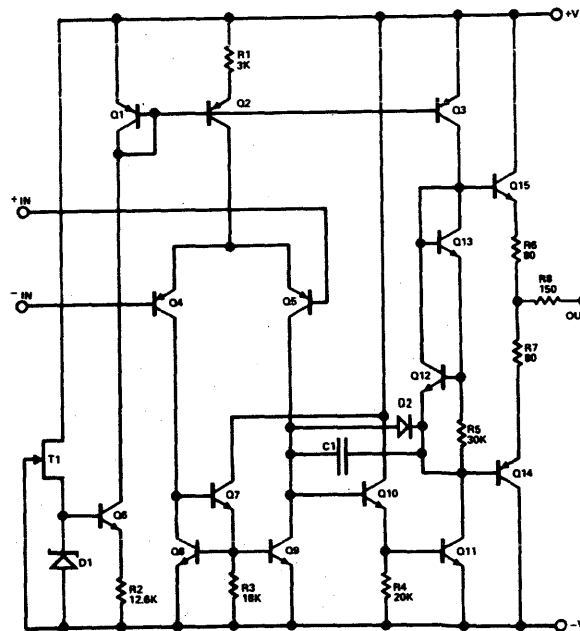
The HA-4156 contains four general purpose operational amplifiers on a monolithic chip. The performance of each amplifier is equal to or better than the 741 type amplifier in all respects. Its superior bandwidth, slew rate and noise characteristics make it an excellent choice for active filter or audio amplifier applications.

The HA-4156-5 is guaranteed over 0°C to +75°C.

### PINOUT



### SCHEMATIC



(%) HA-4156



**ABSOLUTE MAXIMUM RATINGS**

$T_A = +25^\circ\text{C}$ Unless Otherwise Stated		Power Dissipation (Note 3)	880mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	
Differential Input Voltage	$\pm 30.0\text{V}$		
Input Voltage (Note 1)	$\pm 15.0\text{V}$	HA-4156-5	$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$
Output Short Circuit Duration (Note 2)	Indefinite	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

**ELECTRICAL CHARACTERISTICS**

PARAMETER	TEMP.	HA-4156-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>					
Offset Voltage	+25°C		1.0	5.0	mV
	Full		5.0	6.5	mV
Av. Offset Voltage Drift	Full		5		$\mu\text{V}/^\circ\text{C}$
Bias Current	+25°C		60	300	nA
	Full			400	nA
Offset Current	+25°C		30	50	nA
	Full			100	nA
Common Mode Range	Full	$\pm 12$			V
Differential Input Resistance	+25°C		5		M $\Omega$
Input Noise Voltage (f = 1KHz)	+25°C		9		nV/ $\sqrt{\text{Hz}}$
(f = 20Hz to 20kHz)	+25°C		1.4	2.0	$\mu\text{VRMS}$
<b>TRANSFER CHARACTERISTICS</b>					
Large Signal Voltage Gain (Note 4)	+25°C	25K	50K		V/V
	Full	15K			V/V
Common Mode Rejection Ratio (Note 8)	+25°C	80			dB
	Full	74			dB
Channel Separation (Note 5)	+25°C		-108		dB
Small Signal Bandwidth	+25°C	2.8	3.5		MHz
<b>OUTPUT CHARACTERISTICS</b>					
Output Voltage Swing ( $R_L = 10\text{K}$ )	Full	$\pm 12$	$\pm 13.7$		V
( $R_L = 2\text{K}$ )	Full	$\pm 10$	$\pm 12.5$		V
Full Power Bandwidth (Note 4)	+25°C	20	25		KHz
Output Current (Note 6)	Full	$\pm 5$	$\pm 15$		mA
Output Resistance	+25°C		300		$\Omega$
<b>TRANSIENT RESPONSE (Note 7)</b>					
Rise Time	+25°C		75		ns
Overshoot	+25°C		25		%
Slew Rate	+25°C	1.3	$\pm 1.6$		V/ $\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>					
Supply Current ( $I^+$ or $I^-$ )	+25°C			7.0	mA
Power Supply Rejection Ratio (Note 8)	Full	80			dB

- NOTES: 1. For supply voltages less than  $\pm 15\text{V}$ , the absolute maximum input voltage is equal to the supply voltage.  
 2. One amplifier may be shorted to ground indefinitely.  
 3. Derate 5.8mW/ $^\circ\text{C}$  above  $T_A = +25^\circ\text{C}$ .  
 4.  $V_{\text{OUT}} = \pm 10$ ,  $R_L = 2\text{K}$   
 5. Referred to input; f = 10KHz,  $R_S = 1\text{K}$   
 6.  $V_{\text{OUT}} = \pm 10$   
 7. See pulse response characteristics  
 8.  $\Delta V = \pm 5.0\text{V}$

# HA-4600/02/05

## High Performance Quad Operational Amplifier

### FEATURES

- LOW OFFSET VOLTAGE                      0.3mV
- HIGH SLEW RATE                             $\pm 4V/\mu s$
- WIDE BANDWIDTH                          8MHz
- LOW DRIFT                                     $2\mu V/^\circ C$
- FAST SETTLING (0.01%, 10V STEP)    4.2 $\mu s$
- LOW POWER CONSUMPTION               35mW/AMP
- SUPPLY RANGE                               $\pm 5V$  TO  $\pm 20V$

### APPLICATIONS

- HIGH Q, WIDE BAND FILTERS
- INSTRUMENTATION AMPLIFIERS
- AUDIO AMPLIFIERS
- DATA ACQUISITION SYSTEMS
- INTEGRATORS
- ABSOLUTE VALUE CIRCUITS
- TONE DETECTORS

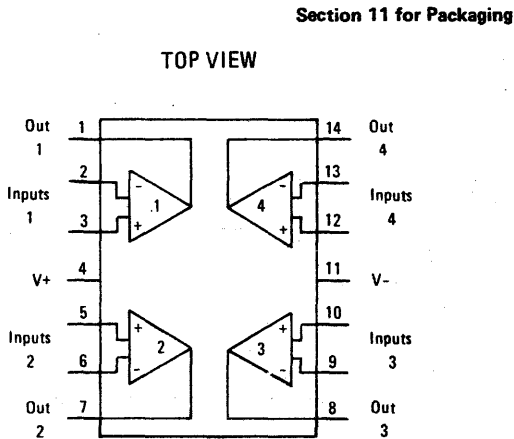
### DESCRIPTION

The HA-4600 series are high performance dielectrically isolated monolithic quad operational amplifiers with superior specifications not previously available in a quad amplifier. These amplifiers offer excellent dynamic performance coupled with low values for offset voltage and drift, input noise voltage and power consumption.

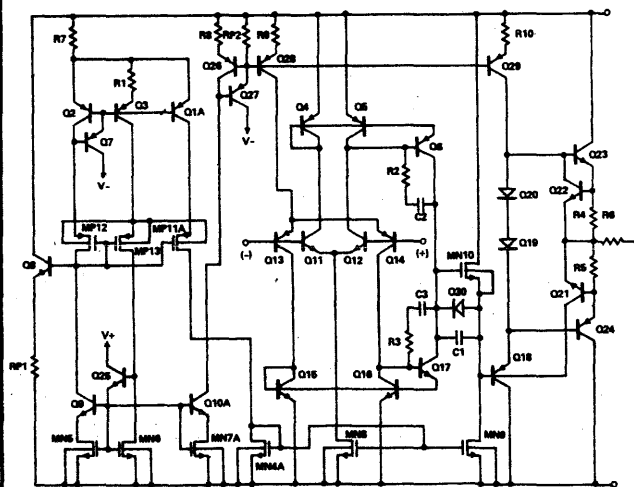
A wide range of applications can be achieved by using the features made available by the HA-4600 series. With wide bandwidth (8MHz), low power (35mW/amp), and internal compensation, these devices are ideally suited for precision active filter designs. For audio applications these amplifiers offer low noise ( $8nV/\sqrt{Hz}$ ) and excellent full power bandwidth (60kHz). The HA-4602/4605 is particularly useful in designs requiring low offset voltage (0.3mV) and drift ( $2\mu V/^\circ C$ ), such as instrumentation and signal conditioning circuits. The high slew rate ( $4V/\mu s$ ) and fast settling time (4.2 $\mu s$  to 0.01%, 10V step) makes these amplifiers useful components in fast, accurate data acquisition systems.

The HA-4600 series are available in 14 pin CERDIP packages which are interchangeable with most other quad op amps. HA-4600/4602-2 is specified from  $-55^\circ C$  to  $+125^\circ C$  and HA-4600/4605-5 is specified over  $0^\circ C$  to  $+75^\circ C$  range.

### PINOUT



### SCHEMATIC



ONE FOURTH ONLY (HA-4600)

Harris Semiconductor LINEAR

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

$T_A = +25^\circ\text{C}$ Unless Otherwise Stated		Power Dissipation (Note 4)	880mW
Voltage Between $V_+$ and $V_-$ Terminals	40.0V	Operating Temperature Range	
Differential Input Voltage	$\pm 7\text{V}$	HA-4600/4602-2	$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$
Input Voltage (Note 2)	$\pm 15.0\text{V}$	HA-4600/4605-5	$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$
Output Short Circuit Duration (Note 3)	Indefinite	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

## ELECTRICAL CHARACTERISTICS $V_+ = +15\text{V}, V_- = -15\text{V}$

PARAMETER	TEMP	HA-4600-2 HA-4600-5			HA-4602-2 HA-4605-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	$+25^\circ\text{C}$		0.3	2.5		3.0	9	mV
	Full			3.0			10	mV
Av. Offset Voltage Drift	Full		2			5		$\mu\text{V}/^\circ\text{C}$
Bias Current	$+25^\circ\text{C}$		130	200		200	400	nA
	Full			325			500	nA
Offset Current	$+25^\circ\text{C}$		30	75		70	150	nA
	Full			125			175	nA
Common Mode Range	Full	$\pm 12$			$\pm 12$			V
Input Noise Voltage ( $f = 1\text{kHz}$ )	$+25^\circ\text{C}$		8			8		$\text{nV}/\sqrt{\text{Hz}}$
Input Resistance			500			500		k $\Omega$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 5)	Full	100K	250K		75K	250K		V/V
Common Mode Rejection Ratio (Note 9)	Full	86			80			dB
Channel Separation (Note 6)	$+25^\circ\text{C}$		-108			-108		dB
Small Signal Bandwidth	$+25^\circ\text{C}$		8			8		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing ( $R_L = 10\text{K}$ ) ( $R_L = 2\text{K}$ )	Full	$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$		V
	Full	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
Full Power Bandwidth (Note 5)	$+25^\circ\text{C}$		60			60		kHz
Output Current (Note 7)	Full	$\pm 10$	$\pm 15$		$\pm 8$	$\pm 15$		mA
Output Resistance	$+25^\circ\text{C}$		200			200		$\Omega$
<b>TRANSIENT RESPONSE (Note 8)</b>								
Rise Time	$+25^\circ\text{C}$		50			50		ns
Overshoot	$+25^\circ\text{C}$		30			30		%
Slew Rate	$+25^\circ\text{C}$	1	$\pm 4$			$\pm 4$		V/ $\mu\text{s}$
Settling Time (Note 10)			4.2			4.2		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	$+25^\circ\text{C}$		4.6	5.5		5.0	7.5	mA
Power Supply Rejection Ratio (Note 9)	Full	86			74			dB

LINEAR

Harris Semiconductor



# HARRIS

## HA-4620/22/25

### Wideband, High Performance Quad Operational Amplifier

#### FEATURES

- Wide Gain Bandwidth Product      70MHz
- High Slew Rate                       $\pm 20V/\mu s$
- Low Offset Voltage                  0.3mV
- Fast Settling (0.01%, 10V Step)    2.5 $\mu s$
- Total Harmonic Distortion      <.01% to 30kHz
- Low Drift                              2 $\mu V/^\circ C$
- Low Power Consumption          35mW/Amp
- Supply Range                         $\pm 5V$  to  $\pm 20V$

#### APPLICATIONS

- High Q Wide Band Filters
- Pulse Amplifiers
- Audio Amplifiers
- Data Acquisition Systems
- Absolute Value Circuits
- Video and R.F. Amplifiers

#### DESCRIPTION

The HA-4620 series are wide band quad operational amplifiers featuring high slew rate, wide bandwidth and fast settling time specifications complemented by low input offset voltage, low drift and input noise voltage.

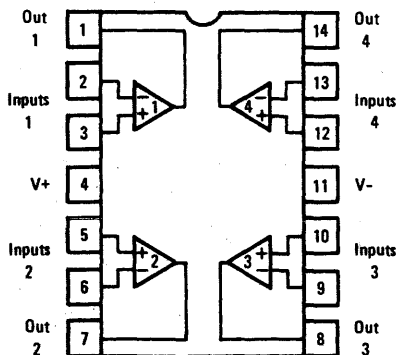
These dielectrically isolated devices are optimized to offer excellent features suitable for applications where a gain of 10 or greater is to be used. The 35mW/amp and a 70MHz gain-bandwidth-product make these monolithic amplifiers valuable components for many active filter circuits. HA-4620 series offers 0.3mV offset voltages and 2 $\mu V/^\circ C$  offset voltage drift for very accurate signal conditioning designs. In high performance audio applications, these amplifiers deliver 260kHz full power bandwidth and 8nV/ $\sqrt{Hz}$  noise voltage. For fast accurate data acquisition systems HA-4620 series offer 20V $\mu s$  slew rate and settling time of 2.5 $\mu s$  to 0.1% 10V step.

HA-4620 series are available in 14 pin CERDIP packages and are interchangeable with most other quad op amps. HA-4625 is also available in chip form. HA-4620/4622-2 is specified from -55 $^\circ C$  to +125 $^\circ C$  and HA-4620/4625-5 is specified over 0 $^\circ C$  to +75 $^\circ C$  range.

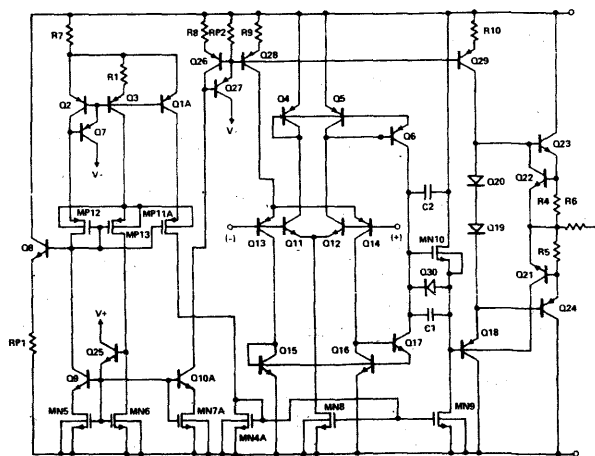
#### PINOUT

Section 11 for Packaging

TOP VIEW



#### SCHEMATIC



ONE FOURTH ONLY (HA-4620)

LINEAR

Harris Semiconductor

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

$T_A = +25^\circ\text{C}$  Unless otherwise stated.

Voltage between V+ and V- Terminals	40.0V
Differential Input Voltage	$\pm 7\text{V}$
Input Voltage (Note 2)	$\pm 15.0\text{V}$
Output Short Circuit Duration (Note 3)	Indefinite

Power Dissipation (Note 4)

880mW

Operating Temperature Range

HA-4620/4622-2

$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$

HA-4620/4625-5

$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$

Storage Temperature Range

$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

## ELECTRICAL CHARACTERISTICS

PARAMETER	TEMP	HA-4620-2 HA-4620-5			HA-4622-2 HA-4625-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C Full		0.3 2.5	2.5 3.0		3.0 9	9 10	mV mV
Av. Offset Voltage Drift	Full		2			5		$\mu\text{V}/^\circ\text{C}$
Bias Current	+25°C Full		130	200 325		200	400 500	nA nA
Offset Current	+25°C Full		30	75 125		70	150 175	nA nA
Common Mode Range	Full	$\pm 12$			$\pm 12$			V
Input Noise Voltage (f = 1kHz)	+25°C		8			8		$\text{nV}/\sqrt{\text{Hz}}$
Input Resistance	+25°C		500			500		k $\Omega$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 5)	Full	100K	250K		75K	250K		V/V
Common Mode Rejection Ratio (Note 6)	Full	86			80			dB
Channel Separation (Note 7)	+25°C		-108			-108		dB
Gain Bandwidth Products (Note 8)	+25°C		70			70		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing ( $R_L = 10\text{K}$ ) ( $R_L = 2\text{K}$ )	Full Full	$\pm 12$ $\pm 10$	$\pm 13$ $\pm 12$		$\pm 12$ $\pm 10$	$\pm 13$ $\pm 12$		V V
Full Power Bandwidth (Note 9)	+25°C		260			260		kHz
Output Current (Note 7)	Full	$\pm 10$	$\pm 15$		$\pm 8$	$\pm 15$		mA
Output Resistance	+25°C		200			200		$\Omega$
<b>TRANSIENT RESPONSE (Note 11)</b>								
Rise Time	+25°C		38	60		38		ns
Overshoot	+25°C		45	60		45		%
Slew Rate	+25°C	$\pm 12$	$\pm 20$		$\pm 12$	$\pm 20$		V/ $\mu\text{s}$
Settling Time (Note 10)			2.5			2.5		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	+25°C		4.6	5.5		5.0	7.5	mA
Power Supply Rejection Ratio (Note 9)	Full	86			74			dB

LINEAR

Harris Semiconductor



## Quad Operational Amplifier

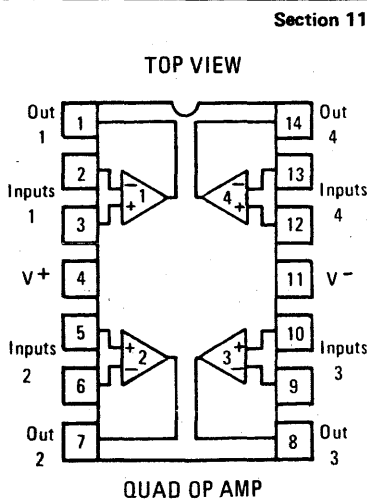
### FEATURES

- SLEW RATE 1.6V/ $\mu$ s (TYP.)
- BANDWIDTH 3.5MHz (TYP.)
- INPUT VOLTAGE NOISE 9nV/ $\sqrt{\text{Hz}}$  (TYP.)
- INPUT OFFSET VOLTAGE 0.5mV (TYP.)
- INPUT BIAS CURRENT 60nA (TYP.)
- SUPPLY RANGE  $\pm 2\text{V TO } \pm 20\text{V}$
- NO CROSSOVER DISTORTION
- STANDARD QUAD PIN-OUT

### APPLICATIONS

- UNIVERSAL ACTIVE FILTERS
- D3 COMMUNICATIONS FILTERS
- AUDIO AMPLIFIERS
- BATTERY-POWERED EQUIPMENT

### PINOUT



### DESCRIPTION

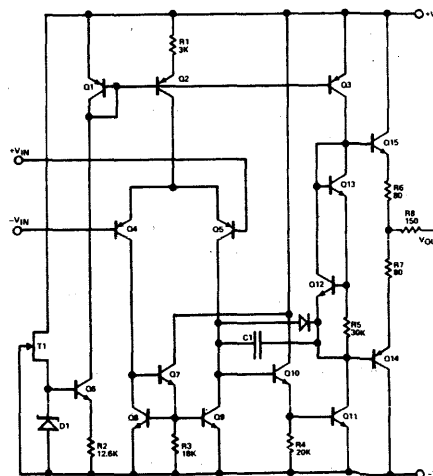
The HA-4741, which contains four amplifiers on a monolithic chip, provides a new measure of performance for general purpose operational amplifiers. Each amplifier in the HA-4741 has operating specifications that equal or exceed those of the 741-type amplifier in all categories of performance.

HA-4741 is well suited to applications requiring accurate signal processing by virtue of its low values of input offset voltage (0.5mV), input bias current (60nA) and input voltage noise (9nV/ $\sqrt{\text{Hz}}$  at 1kHz). 3.5MHz bandwidth, coupled with high open-loop gain, allow the HA-4741 to be used in designs requiring amplification of wide band signals, such as audio amplifiers. Audio application is further enhanced by the HA-4741's negligible output crossover distortion. These excellent dynamic characteristics also make the HA-4741 ideal for a wide range of active filter designs. Performance integrity of multi-channel designs is assured by a high level of amplifier-to-amplifier isolation (108dB at 1kHz).

A wide range of supply voltages ( $\pm 2\text{V to } \pm 20\text{V}$ ) can be used to power the HA-4741, making it compatible with almost any system including battery-powered equipment.

The HA-4741 has guaranteed operation over  $-55^{\circ}\text{C to } +125^{\circ}\text{C}$  and can be furnished to meet MIL-STD-883 (HA-4741-8). The HA-4741-5 is guaranteed over  $0^{\circ}\text{C to } +75^{\circ}\text{C}$  and is available in ceramic and plastic dual-in-line packages and in dice form.

### SCHEMATIC



( $\frac{1}{4}$ ) HA-4741

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

$T_A = +25^\circ\text{C}$ Unless Otherwise Stated		Power Dissipation For Epoxy Package. (Note 3)	880mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	
Differential Input Voltage	$\pm 30.0\text{V}$	HA-4741-2	$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$
Input Voltage (Note 1)	$\pm 15.0\text{V}$	HA-4741-5	$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$
Output Short Circuit Duration (Note 2)	Indefinite	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

## ELECTRICAL CHARACTERISTICS

PARAMETER	TEMP.	HA-4741-2 -55°C to +125°C			HA-4741-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		0.5	3.0		1.0	5.0	mV
	Full		4.0	5.0		5.0	6.5	mV
Av. Offset Voltage Drift	Full		5			5		$\mu\text{V}/^\circ\text{C}$
Bias Current	+25°C		60	200		60	300	nA
	Full			325			400	nA
Offset Current	+25°C		15	30		30	50	nA
	Full			75			100	nA
Common Mode Range	Full	$\pm 12$			$\pm 12$			V
Differential Input Resistance	+25°C		5			5		M $\Omega$
Input Noise Voltage (f = 1KHz)	+25°C		9			9		$\text{nV}/\sqrt{\text{Hz}}$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 4)	+25°C	50K	100K		25K	50K		V/V
	Full	25K			15K			V/V
Common Mode Rejection Ratio (Note 8)	+25°C	80			80			dB
	Full	74			74			dB
Channel Separation (Note 5)	+25°C	90	-108		90	-108		dB
Small Signal Bandwidth	+25°C	2.5	3.5		2.5	3.5		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing ( $R_L = 10\text{K}$ )	Full	$\pm 12$	$\pm 13.7$		$\pm 12$	$\pm 13.7$		V
( $R_L = 2\text{K}$ )	Full	$\pm 10$	$\pm 12.5$		$\pm 10$	$\pm 12.5$		V
Full Power Bandwidth (Notes 4 & 9)	+25°C	14	25		14	25		kHz
Output Current (Note 6)	Full	$\pm 5$	$\pm 15$		$\pm 5$	$\pm 15$		mA
Output Resistance	+25°C		300			300		$\Omega$
<b>TRANSIENT RESPONSE (Notes 7 &amp; 10)</b>								
Rise Time (Note 11)	+25°C		75	140		75	140	ns
Overshoot (Note 11)	+25°C		25	40		25	40	%
Slew Rate (Note 12)	+25°C		$\pm 1.6$			$\pm 1.6$		V/ $\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	+25°C			5.0			7.0	mA
Power Supply Rejection Ratio (Note 8)	Full	80			80			dB

- NOTES:
- For supply voltages less than  $\pm 15\text{V}$ , the absolute maximum input voltage is equal to the supply voltage.
  - One amplifier may be shorted to ground indefinitely.
  - Derate 5.8mW/ $^\circ\text{C}$  above  $T_A = +25^\circ\text{C}$ .
  - $V_{\text{OUT}} = \pm 10$ ,  $R_L = 2\text{K}$
  - Referred to input; f = 10KHz,  $R_S = 1\text{K}$
  - $V_{\text{OUT}} = \pm 10$
  - See pulse response characteristics
  - $\Delta V = \pm 5.0\text{V}$
  - Full power bandwidth guaranteed based upon slew rate measurement FPBW = S.R./ $2\pi V_{\text{peak}}$
  - $R_L = 2\text{K}$ ,  $C_L = 50\text{pf}$ .
  - $V_{\text{OUT}} = \pm 200\text{mV}$
  - $V_{\text{OUT}} = \pm 5\text{V}$

LINEAR

Harris Semiconductor



# HARRIS

## HA-5062 Series

### Low Power JFET Input Dual Operational Amplifiers

*Preliminary*

#### FEATURES

- HIGH INPUT IMPEDANCE  $10^{12}\Omega$
- LOW INPUT BIAS CURRENT 200pA
- LOW INPUT OFFSET CURRENT 100pA
- VERY LOW POWER CONSUMPTION  
TYPICAL SUPPLY CURRENT  $200\mu\text{A}$
- INTERNAL FREQUENCY COMPENSATION
- HIGH SLEW RATE  $4\text{V}/\mu\text{s}$
- PIN COMPATIBLE WITH LM1458
- DIRECT REPLACEMENT FOR TL062

#### DESCRIPTION

The HARRIS HA-5062 operational amplifiers are a series of dual monolithic JFET-input amplifiers featuring low input bias and offset currents, high input impedance and very low power operation. In addition to being a direct replacement for the TL062 series, the HA-5062 series offers improved performance with a minimum open loop gain 20K V/V and a slew rate of  $4\text{V}/\mu\text{s}$ .

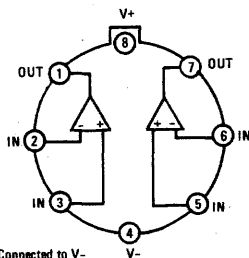
This improved performance is a result of the HARRIS FET/Bipolar technology and makes the HA-5062 series of amplifiers ideally suited for applications in industrial control, communication, and battery powered instrumentation equipment.

The HA-5062-2 is characterized for operation over the full military temperature range of  $-55^\circ\text{C}$  to  $+125^\circ\text{C}$ . The HA-5062A-5, HA-5062B-5 and HA-5062-5 are all characterized over the commercial temperature range of  $0^\circ\text{C}$  to  $+75^\circ\text{C}$ .

#### APPLICATIONS

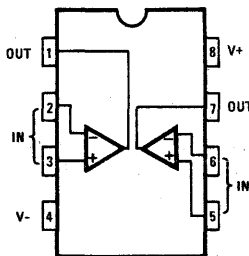
- ACTIVE FILTERS
- INSTRUMENTATION AMPLIFIERS
- AUDIO AMPLIFIERS
- BATTERY OPERATED EQUIPMENT
- SIGNAL CONDITIONING

#### PINOUT

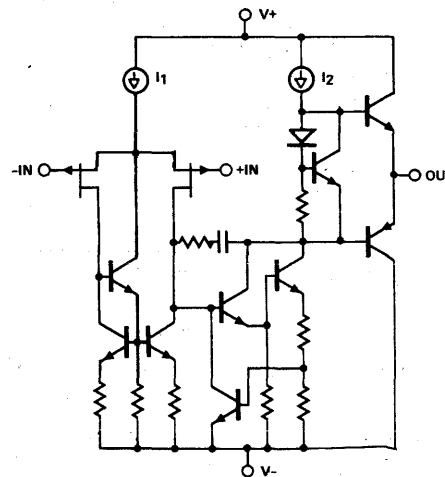


NOTE: Case Connected to V-

TOP VIEWS



#### SIMPLIFIED SCHEMATIC



(ONE HALF ONLY)

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Voltage Between V+ and V- Terminals	± 20V
Differential Input Voltage	± 40V
Input Voltage (Note 2)	±15.0V
Output Short Circuit Duration	Indefinite

Power Dissipation

600mW\*

Operating Temperature Range:

HA-5062-2

$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$

HA-5062-5

$0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$

Storage Temperature Range

$-65^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$

\* To-99 Derate by 6.75mW/°C above +85°C

Dip Derate by 5.57mW/°C above +65°C

## ELECTRICAL CHARACTERISTICS

V+ = 15V V- = 15V

Parameters are guaranteed at indicated ambient temperature after warm-up.

PARAMETER	TEMP.	HA-5062-2 -55°C to +125°C			HA-5062A-5 0°C to 75°C			HA-5062B-5 0°C to 75°C			HA-5062-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage (Note 3)	+25°C		3	6		3	6		2	3		3	15	mV
	Full			9			7.5			5			20	mV
Av. Offset Voltage Drift	Full		10			10			10			10		μV/°C
Bias Current	+25°C		30	200		30	200		30	200		30	400	pA
	Full			50			7			7			10	nA
Offset Current	+25°C		5	100		5	100		5	100		5	200	pA
	Full			20			3			3			5	nA
Common Mode Range	Full	±10	±12		±10	±12		±10	±12		±10	±12		V
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		MΩ
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 4)	+25°C	20K	25K		20K	25K		20K	25K		10K	25K		V/V
	Full	10K			15K			15K			5K			V/V
Common Mode Rejection Ratio (Note 5)	Full	80	86		80	86		80	86		70	76		dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 6)	+25°C	±10	±12		±10	±12		±10	±12		±10	±12		V
	Full	±10			±10			±10			±10			V
Unity Gain Bandwidth (Note 6)	+25°C		1			1			1			1		MHz
Full Power Bandwidth (Note 7)	+25°C		63			63			63			63		KHz
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 8)	+25°C		80			80			80			80		nsec
Overshoot (Note 8)	+25°C		10			10			10			10		%
Slew Rate (Note 9)	+25°C		4			4			4			4		V/μs
Settling Time (Note 10)	+25°C		3.5			3.5			3.5			3.5		μsec
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current (Note 11)	+25°C			0.4			0.4			0.4			0.5	mA
Power Supply Rejection Ratio (Note 12)	Full	80	95		80	95		80	95		70	95		dB

### NOTES:

- Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- For supply voltages less than ±15V, the absolute maximum input voltage is equal to the supply voltage.
- R<sub>S</sub> = 50Ω.
- R<sub>L</sub> ≥ 10KΩ, V<sub>O</sub> = ±10V.
- ΔV<sub>IN</sub> = ±10V.
- R<sub>L</sub> = 10KΩ.
- R<sub>L</sub> = 10KΩ; Full power bandwidth guaranteed based on slew rate measurement using  $FPBW = \frac{SLEW\ RATE}{2\pi V_{PEAK}}$ .
- V<sub>IN</sub> = 50mV, C<sub>L</sub> = 50pF, R<sub>L</sub> = 10KΩ.
- V<sub>IN</sub> = 10V, C<sub>L</sub> = 50pF, R<sub>L</sub> = 10KΩ.
- Settling time is measured to 0.1% of final value for a 10 volt output step and A<sub>V</sub> = -1.
- No load, No signal.
- V<sub>SUPP</sub> = ±5V.D.C. to ±15 V.D.C.



# HARRIS

# HA-5064 Series

## Low Power, JFET Input Quad Operational Amplifiers

JULY 1982

*Preliminary*

### FEATURES

- LOW INPUT BIAS CURRENT            100pA
- LOW POWER DISSIPATION            24mW/Pkg.
- FAST SLEWING                        4V/ $\mu$ s
- LOW VIO DRIFT                        10 $\mu$ V/ $^{\circ}$ C
- HIGH INPUT IMPEDANCE            10<sup>12</sup> $\Omega$
- GOOD CHANNEL SEPARATION        120dB
- POWER SUPPLY RANGE             $\pm$ 5V TO  $\pm$ 20V

### DESCRIPTION

The HARRIS HA-5064 series JFET input monolithic, quad operational amplifiers feature very low power requirements coupled with excellent AC and DC characteristics. Maximum power dissipation of 24 mW/package is achieved by using complementary design, process, and layout techniques.

A 4V/ $\mu$ s slew rate coupled with 1MHz gain-bandwidth makes these devices most suitable for active filter and signal conditioning designs. The HA-5064 series is ideally suited for those applications demanding low power and high density without compromising other performance characteristics. High input impedance and low drift also makes the HA-5064 series useful as instrumentation amplifiers.

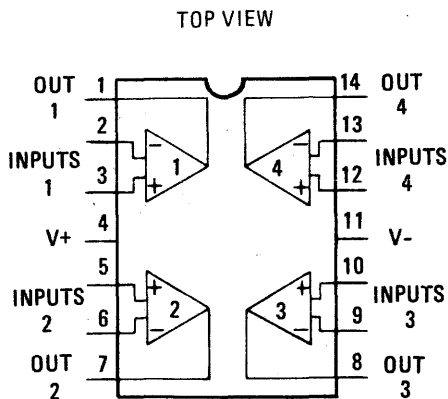
The HA-5064 is packaged in a 14-pin DIP and is pin compatible with most other quad operational amplifiers. The HA-5064-2 is specified for -55 $^{\circ}$ C to +125 $^{\circ}$ C operation while the HA-5064 A-5/HA-5064B-5/HA-5064-5 are specified over the 0 $^{\circ}$ C to +75 $^{\circ}$ C range.

### APPLICATIONS

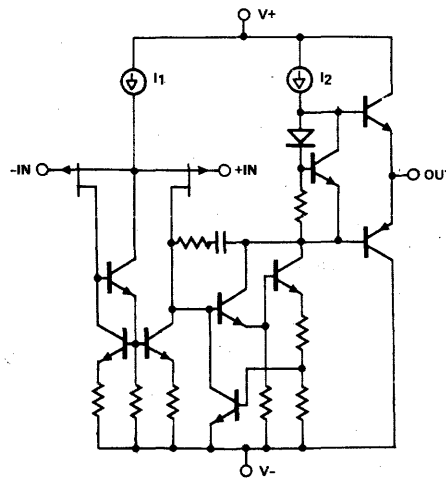
WHERE DENSITY AND POWER REQUIREMENTS ARE DEMANDING:

- ACTIVE FILTERS
- SIGNAL CONDITIONING
- SIGNAL GENERATION
- INSTRUMENTATION AMPLIFIERS

### PINOUT



### SIMPLIFIED SCHEMATIC



ONE-FOURTH ONLY

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# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V-	40V
Differential Input Voltage (Note 2)	±30V
Output Current (Note 3)	Continuous
Internal Power Dissipation (Note 4)	500mW
Storage Temperature Range	-65°C to +150°C

## ELECTRICAL CHARACTERISTICS

V+ = 15VDC; V- = -15VDC

PARAMETER	TEMP	HA-5064-2 -55°C to +125°C			HA-5064A-5 0°C to +75°C			HA-5064B-5 0°C to +75°C			HA-5064-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	+25°C		2	6		2	6		2	3			15	mV
	Full			9			7.5			5			20	mV
Offset Voltage Average Drift	Full		10			10			10			20		μV/°C
Bias Current	+25°C			200			200			200			400	pA
	Full			50			7			7			10	nA
Offset Current	+25°C			100			100			100			200	pA
	Full			20			3			3			5	nA
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω
Common Mode Range	Full	±10			±10			±10			±10			V
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 5)	+25°C	20K	25K		20K	25K		20K	25K		10K	25K		V/V
	Full	10K			15K			15K			5K			V/V
Common Mode Rejection Ratio (Note 6)	Full	80			80			80			70			dB
Gain Bandwidth	+25°C		1			1			1			1		MHz
Channel Separation (Note 7)	+25°C		120			120			120			120		dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 8)	+25°C	±10	±12		±10	±12		±10	±12		±10	±12		V
	Full	±10			±10			±10			±10			V
Output Current (Note 9)	Full	±1			±1			±1			±1			mA
Full Power Bandwidth (Note 10)	+25°C		63			63			63			63		kHz
Output Resistance (Note 11)	+25°C		300			300			300			300		Ω
<b>TRANSIENT RESPONSE (Note 12)</b>														
Rise Time (10% TO 90%)	+25°C		80			80			80			80		nsec
Slew Rate	+25°C	2	4		2	4		2	4		2	4		V/μsec
Settling Time (Note 13)	+25°C		3.5			3.5			3.5			3.5		μsec
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current	+25°C			.8		.8			.8			1		mA
P. S. R. R. (Note 14)	Full	80			80			80			70			dB

LINEAR

Harris Semiconductor

*Preliminary*

*JFET Input  
Dual Operational Amplifiers*

### FEATURES

- HIGH INPUT IMPEDANCE 10<sup>12</sup>Ω
- LOW INPUT BIAS CURRENT 200pA
- LOW INPUT OFFSET CURRENT 100pA
- LOW POWER CONSUMPTION 3.5mA
- TYPICAL SUPPLY CURRENT 15V/μs
- HIGH SLEW RATE
- PIN COMPATIBLE WITH LM1458
- DIRECT REPLACEMENT FOR TL082

### DESCRIPTION

The HARRIS HA-5082 operational amplifiers are a series of dual monolithic JFET-input amplifiers featuring low input bias and offset currents, high input impedance and, high slew rate. In addition to being a direct replacement for the TL082 series, the HA-5082 series offers improved performance with an input offset voltage of 2mV, a slew rate of 15V/μs, and bandwidths of 4MHz.

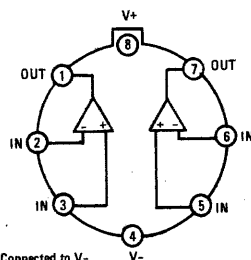
This improved performance is a result of the HARRIS FET/Bipolar technology and makes the HA-5082 series of amplifiers ideally suited for applications in industrial control, communication, and computer peripheral equipment.

The HA-5082-2 is characterized for operation over the full military temperature range of -55°C to +125°C. The HA-5082A-5, HA-5082B-5 and HA-5082-5 are all characterized over the commercial temperature range of 0°C to +75°C.

### APPLICATIONS

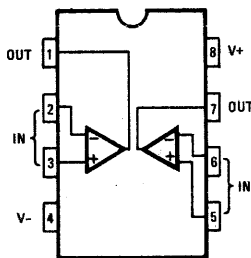
- ACTIVE FILTERS
- INSTRUMENTATION AMPLIFIERS
- AUDIO AMPLIFIERS
- SIGNAL CONDITIONING

### PINOUT

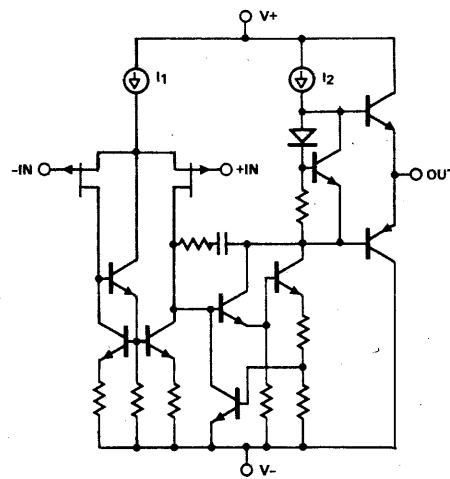


NOTE: Case Connected to V-

TOP VIEWS



### SIMPLIFIED SCHEMATIC



(ONE HALF ONLY)

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Voltage Between V+ and V- Terminals	±20V	Power Dissipation	600mW*
Differential Input Voltage	±40V	Operating Temperature Range:	HA-5082-2      -55°C ≤ T <sub>A</sub> ≤ +125°C
Input Voltage (Note 2)	±15.0V		HA-5082-5      0°C ≤ T <sub>A</sub> ≤ +75°C
Output Short Circuit Duration	Indefinite	Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C

\* To-99 Derate by 6.75mW/°C above +85°C  
Dip Derate by 5.57mW/°C above +65°C

## ELECTRICAL CHARACTERISTICS

V+ = 15V V- = -15V

Parameters are guaranteed at indicated ambient temperature after warm-up.

PARAMETER	TEMP.	HA-5082-2 -55°C to +125°C			HA-5082A-5 0°C to 75°C			HA-5082B-5 0°C to 75°C			HA-5082-5 0°C to +75°C			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage (Note 3)	+25°C		3	5		3	5		2		5	15	mV	
	Full			8			7		4			20	mV	
Av. Offset Voltage Drift	Full		10			10			10		10		μV/°C	
Bias Current	+25°C		30	200		30	200		30	200		30	400	pA
	Full			50			8		4				10	nA
Offset Current	+25°C		5	100		5	100		5	100		5	200	pA
	Full			20			4		2				5	nA
Common Mode Range	Full	±10	±12		±10	±12		±10	±12		±10	±12		V
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		MΩ
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 4)	+25°C	50K	200K		50K	200K		50K	200K		25K	200K		V/V
	Full	15K			25K			25K			15K			V/V
Common Mode Rejection Ratio (Note 5)	+25°C	80	86		80	86		80	86		70	76		dB
Unity Gain Bandwidth	+25°C		4			4			4			4		MHz
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 6)	+25°C	±10	±12		±10	±12		±10	±12		±10	±12		V
	Full	±10			±10			±10			±10			V
Output Current (Note 7)	Full	±5			±5			±5			±5			mA
Full Power Bandwidth (Note 8)	+25°C		240			240			240			240		KHz
<b>TRANSIENT RESPONSE</b>														
Rise Time (Note 9)	+25°C		60			60			60			60		nsec
Overshoot (Note 9)	+25°C		10			10			10			10		%
Slew Rate (Note 10)	+25°C		15			15			15			15		V/μs
Settling Time (Note 11)	+25°C		2			2			2			2		μsec
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current (Note 12)	+25°C		3.5	5.6		3.5	5.6		3.5	5.6		3.5	5.6	mA
Power Supply Rejection Ratio (Note 13)	+25°C	80	86		80	86		80	86		70	76		dB

### NOTES:

- Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- For supply voltages less than ±15V, the absolute maximum input voltage is equal to the supply voltage.
- R<sub>S</sub> = 50Ω.
- R<sub>L</sub> ≥ 2KΩ, V<sub>O</sub> = ±10V.
- ΔV<sub>IN</sub> = ±10V.
- R<sub>L</sub> = 2KΩ.
- V<sub>OUT</sub> = ±10V
- R<sub>L</sub> = 2K; Full power bandwidth guaranteed based on slew rate measurement using  $FPBW = \frac{SLEW\ RATE}{2\pi V_{PEAK}}$
- V<sub>IN</sub> = 50mV, C<sub>L</sub> = 100pF, R<sub>L</sub> = 2KΩ.
- V<sub>IN</sub> = 10V, C<sub>L</sub> = 100pF, R<sub>L</sub> = 2KΩ.
- Settling time is measured to 0.1% of final value for a 10 volt output step and A<sub>V</sub> = -1.
- No load, No signal.
- V<sub>SUPP</sub> = ±5V.D.C. to ±15 V.D.C.





# HARRIS

# HA-5084 Series

## JFET Input Quad Operational Amplifiers

### Preliminary

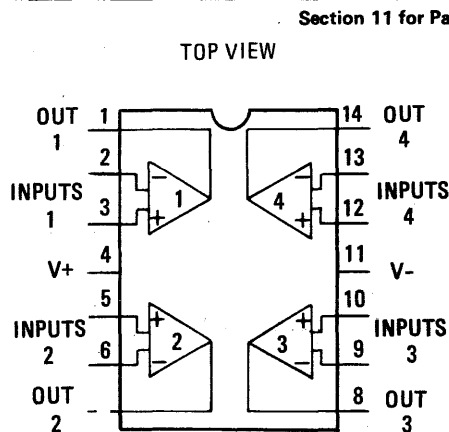
### FEATURES

- LOW INPUT BIAS CURRENT 200pA
- HIGH SLEW RATE 15V/ $\mu$ s
- WIDE BANDWIDTH 4MHz
- LOW DRIFT 10 $\mu$ V/ $^{\circ}$ C
- HIGH INPUT IMPEDANCE 10<sup>12</sup> $\Omega$
- LOW SUPPLY CURRENT 7.2mA
- SUPPLY RANGE  $\pm$ 5V TO  $\pm$ 20V

### APPLICATIONS

- HIGH Q, WIDEBAND FILTERS
- INTEGRATORS
- TONE DETECTORS
- SAMPLE/HOLD CIRCUITS
- DATA ACQUISITION SYSTEMS
- ABSOLUTE VALUE CIRCUITS

### PINOUT



### DESCRIPTION

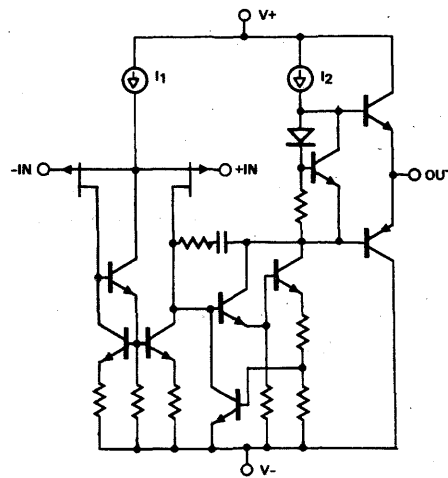
The Harris HA-5084 is a JFET input, monolithic, quad operational amplifier featuring low input bias and offset currents, high input impedance, and high slew rate. Manufactured using FET/Bipolar technology coupled with advanced layout considerations, these devices also feature excellent channel separation and offset voltage drift specifications.

High slew rate (15V/ $\mu$ s) coupled with excellent input bias (30pA) and offset current (3pA) make the HA-5084 ideally suited for high speed analog designs such as integrators, fast D/A converters, and sample-and-hold circuits.

The HA-5084 is available in ceramic and plastic 14 pin DIP's and is pin compatible with the LM324, LM348, and MC3403 quad operational amplifier pinout.

The HA-5084-2 is specified from -55 $^{\circ}$ C to +125 $^{\circ}$ C while the HA-5084-5 operates from 0 $^{\circ}$ C to +75 $^{\circ}$ C.

### SIMPLIFIED SCHEMATIC



(ONE FOURTH ONLY)

LINEAR

Harris Semiconductor

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V-	40V
Differential Input Voltage (Note 2)	±40V
Output Current (Note 3)	Full Short Circuit Protection
Internal Power Dissipation (Note 4)	500mW
Storage Temperature Range	-65°C to +150°C

## ELECTRICAL CHARACTERISTICS

V+ = 15VDC; V- = -15VDC

PARAMETER	TEMP	HA-5084-2 -55°C to +125°C			HA-5084A-5 0°C to +75°C			HA-5084B-5 0°C to +75°C			HA-5084-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	+25°C			5			5			2			15	mV
	Full			8			7			4			20	mV
Offset Voltage Average Drift	Full		8.3			8.3			8.3			8.3		μV/°C
Bias Current	+25°C			200			200			200			400	pA
	Full			50			8			4			10	nA
Offset Current	+25°C			100			100			100			200	pA
	Full			20			4			2			5	nA
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω
Common Mode Range	Full	±10			±10			±10			±10			V
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 5)	+25°C	25K			50K			50K			25K			V/V
	Full	15K			25K			25K			15K			V/V
Common Mode Rejection Ratio (Note 6)	Full	70			80			80			70			dB
Unity Gain Bandwidth	+25°C		4			4			4			4		MHz
Channel Separation (Note 7)			-120			-120			-120			-120		dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing (Note 8)	+25°C	10			10			10			10			V
	Full	10			10			10			10			V
Output Current (Note 9)	Full	±5			±5			±5			±5			mA
Full Power Bandwidth (Note 10)	+25°C		240			240			240			240		kHz
Output Resistance (Note 11)	+25°C		300			300			300			300		Ω
<b>TRANSIENT RESPONSE (Note 12)</b>														
Rise Time	+25°C		60			60			60			60		nsec
Slew Rate	+25°C		15			15			15			15		V/μsec
Settling Time (Note 13)	+25°C		2			2			2			2		μsec
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current	+25°C		7.2	11		7.2	11		7.2	11		7.2	12	mA
P. S. R. R. (Note 14)	Full	80			80			80			70			dB

LINEAR

Harris Semiconductor

### FEATURES

- LOW INPUT OFFSET VOLTAGE . . . . . 0.5mV
- LOW OFFSET DRIFT . . . . . 5 $\mu$ V/°C
- LOW INPUT BIAS CURRENT . . . . . 50pA
- LARGE VOLTAGE GAIN . . . . . 150K V/V
- WIDE BANDWIDTH . . . . . 18MHz
- HIGH SLEW RATE . . . . . 8V/ $\mu$ sec
- FAST LARGE SIGNAL SETTLING TIME: 1.7 $\mu$ sec

### APPLICATIONS

- PRECISION, HIGH SPEED, DATA ACQUISITION SYSTEMS
- PRECISION SIGNAL GENERATION
- PULSE AMPLIFICATION

### GENERAL DESCRIPTION

The HA-5100/5105 are monolithic wideband operational amplifiers manufactured with FET/Bipolar technologies and dielectric isolation. Precision laser trimming of the input stage complements the amplifier high frequency capabilities with excellent input characteristics.

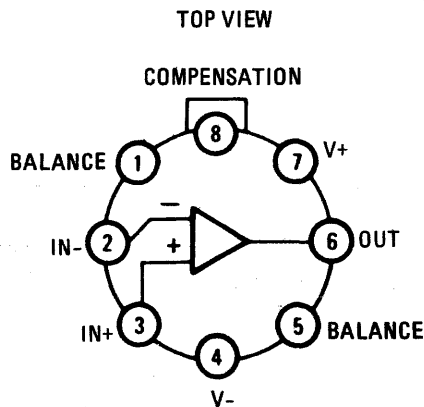
The HA-5100/5105 offer a number of important advantages over similar FET input op amps from other manufacturers. In addition to superior bandwidth and settling characteristics the Harris devices have quite constant slew rate, bandwidth, and settling characteristics over the operating range. This provides the user predictable performance in applications where settling time, full power bandwidth, closed loop bandwidth, or phase shift is critical. The slewing waveform is symmetrical to provide reduced distortion. Note also that Harris specifies all parameters at ambient (rather than junction) temperature to provide the designer meaningful data to predict actual operating performance.

Complementing HA-5100/5105's predictable and excellent dynamic characteristics are very low input offset voltage, very low input bias current, and extremely high input impedance. This ideal combination of features make these amplifiers most suitable for precision, high speed, data acquisition system designs and for a wide variety of signal conditioning applications.\*

\* -2 denotes a range of -55°C to +125°C and -5 denotes a 0°C to +75°C range.

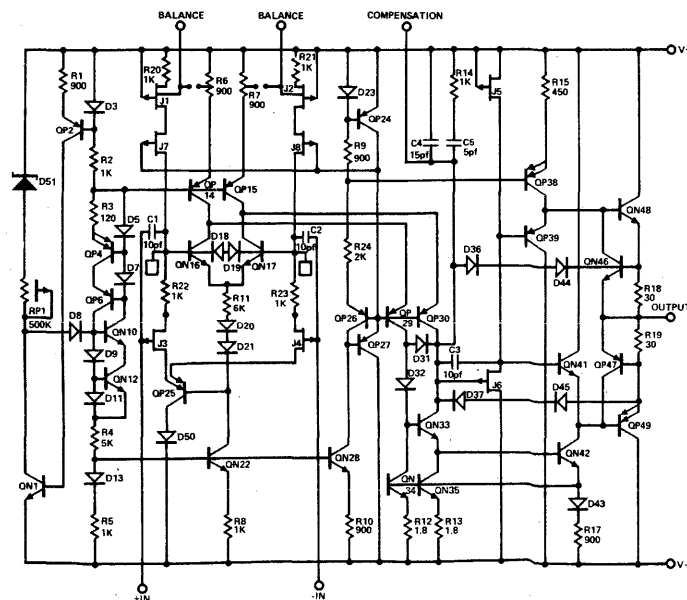
### PINOUT

Section 11 for Packaging



CASE  
CONNECTED  
TO V-

### SCHEMATIC DIAGRAM



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Voltage Between V+ and V-	40V
Differential Input Voltage	±40V
Peak Output Current	Full Short Circuit Protection
Internal Power Dissipation (Note 2)	510mW
Storage Temperature Range	-65°C to +150°C

## ELECTRICAL CHARACTERISTICS

PARAMETER	TEMP	HA-5100-2 -55°C to +125°C			HA-5100-5 0°C to +75°C			HA-5105-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C		0.5	1.0		0.5	1.0		0.5	1.5	mV
	Full		0.50	2.0		0.50	2.0		0.75	3.5	mV
Offset Voltage Average Drift	Full		5			10			15		μV/°C
Bias Current	+25°C		20	50		20	50		50	100	pA
	Full		5	10		10	10		10	20	nA
Offset Current	+25°C		2	10		2	10		5	50	pA
	Full		2	5		2	5		5	10	nA
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω
Common Mode Range	Full	±10	±11		±10	±11		±10	±10.5		V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Note 3)	+25°C	75K	150K		75K	150K		50K	100K		V/V
	Full	60K	100K		60K	100K		40K	80K		V/V
Common Mode Rejection Ratio (Note 4)	Full	80	86		80	86		80	86		dB
Gain Bandwidth Product at A <sub>v</sub> = 10	Full		18			18			18		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 5)	+25°C	±12	±13		±12	±13		±11	±12		V
	Full	±12	±13		±12	±13		±11	±12		V
Short Circuit Output Current (Note 6)	Full	±10	±15		±10	±15		±8	±15		mA
Full Power Bandwidth (Note 7)	+25°C	90	150		90	150		75	125		kHz
Output Resistance (Note 8)	+25°C		30			30			40		Ω
<b>TRANSIENT RESPONSE (Note 9)</b>											
Rise Time	+25°C		15	35		15	35		20		nsec.
Slew Rate	+25°C	6	8		6	8		5	8		V/μsec
Settling Time (Note 10)	+25°C		1.7			1.7			2.0		μsec
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	Full		5	7		5	7		6	8	mA
P.S.R.R. (Note 11)	Full	80	86		80	86		80	86		dB

LINEAR

Harris Semiconductor

## FEATURES

- WIDE GAIN BANDWIDTH . . . . . 60MHz
- HIGH SLEW RATE . . . . . 50V/ $\mu$ s
- SETTLING TIME . . . . . 850ns
- POWER BANDWIDTH . . . . . 800KHz
- OFFSET VOLTAGE . . . . . 0.5mV
- BIAS CURRENT . . . . . 50pA

## APPLICATIONS

- VIDEO AND RF AMPLIFIERS
- DATA ACQUISITION
- PULSE AMPLIFIERS
- PRECISION SIGNAL GENERATION

## DESCRIPTION

HA-5110/5115 are wideband, uncompensated, operational amplifiers manufactured with FET/Bipolar technologies and dielectric isolation. These monolithic amplifiers feature superior high frequency capabilities further enhanced by precision laser trimming of the input stage to provide excellent input characteristics. These devices are controlled at closed loop gains greater than 10 without compensation.

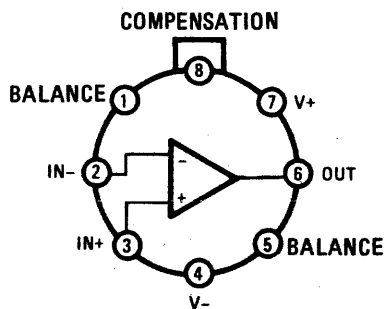
With excellent dynamic and input characteristics, HA-5110/5115 are well suited for many wideband, pulse, and video applications. These amplifiers are ideal components for video and RF circuitry requiring up to 60MHz gain-bandwidth-product and 800KHz power bandwidth. 50V/ $\mu$ s slew rate and 850ns settling time make these devices useful in pulse amplification and data acquisition designs. HA-5110/5115's 0.5mV offset voltage, 10pA offset current, and extremely high impedance coupled with excellent AC parameters make these amplifiers ideal selections for accurate signal conditioning designs. For applications requiring less critical input characteristics, HA-5115 is available in untrimmed form.

HA-5110/5115 are available in metal can (TO-99) packages. Suffix -2 denotes a range to -55°C to +125°C and -5 denotes a 0°C to +75°C range.

## PINOUT

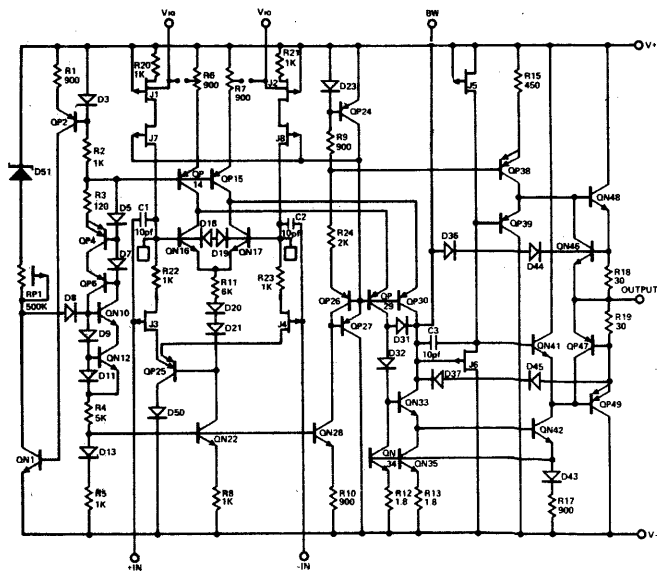
TOP VIEW

Section 11 for Packaging



CASE CONNECTED TO V-

## SCHEMATIC



**ABSOLUTE MAXIMUM RATINGS**

Voltage Between V <sup>+</sup> and V <sup>-</sup>	40V	Internal Power Dissipation (Note 2)	510mW
Differential Input Voltage	±40V	Storage Temperature Range	-65°C to +150°C
Peak Output Current	Full Short Circuit Protection		

**ELECTRICAL CHARACTERISTICS**

V<sup>+</sup> = 15VDC; V<sup>-</sup> = -15VDC  
 Parameters are guaranteed at indicated ambient temperature after warm-up.

PARAMETER	TEMP	HA-5110-2 -55°C to +125°C			HA-5110-5 0°C to +75°C			HA-5115-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C		0.5	1.0		0.5	1.0		0.5	1.5	mV
	Full		0.50	2.0		0.50	2.0		0.75	3.5	mV
Offset Voltage Average Drift	Full		5			10			15		μV/°C
Bias Current	+25°C		20	50		20	50		50	100	pA
	Full		5	10		10	10		10	20	nA
Offset Current	+25°C		2	10		2	10		5	50	pA
	Full		2	5		2	5		5	10	nA
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω
Common Mode Range	Full	±10	±11		±10	±11		±10	±10.5		V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Note 3)	+25°C	75K	150K		75K	150K		50K	100K		V/V
	Full	60K	100K		60K	100K		40K	80K		V/V
Common Mode Rejection Ratio (Note 4)	Full	80	86		80	86		80	86		dB
Gain Bandwidth Product (A <sub>V</sub> = 10)	Full		60			60			50		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 5)	+25°C	±12	±13		±12	±13		±11	±12		V
	Full	±12	±13		±12	±13		±11	±12		V
Output Current (Note 6)	+25°C	±10	±15		±10	±15		±8	±15		mA
Full Power Bandwidth (Note 7)	+25°C	550	625		550	625		550	625		kHz
Output Resistance (Note 8)	+25°C		30			30			40		Ω
<b>TRANSIENT RESPONSE (Note 9)</b>											
Rise Time (A <sub>V</sub> = 10)	+25°C		20			20			20		nsec
Slew Rate (A <sub>V</sub> = 10)	+25°C	35	50		35	50		35	40		V/μsec
Settling Time (Note 10)	+25°C		.85			.85			1.0		μsec
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	Full		5	7		5	7		6	8	mA
Power Supply Rejection Ratio (Note 11)	+25°C	80	94		80	94		80	94		dB

**NOTES:**

- Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- Derate at 6.8mW/°C for operation at ambient temperatures above +75°C.
- V<sub>OUT</sub> = ±10V. R<sub>L</sub> = 2K
- V<sub>CM</sub> = ±10 V.D.C.
- R<sub>L</sub> = 10K
- V<sub>OUT</sub> = 0V
- R<sub>L</sub> = 2K; Full power bandwidth guaranteed, based on slew rate measurement using  $FPBW = \frac{SLEW\ RATE.}{2\pi V_{PEAK}}$
- Output resistance measured under open loop conditions.
- Refer to Test Circuits section of the data sheet.
- Settling Time is measured to 0.1% of final value for a 10 volt output step and A<sub>V</sub> = -10.
- V<sub>SUPP</sub> = ±10 V.D.C. to ±20 V.D.C.

### FEATURES

- LOW OFFSET VOLTAGE 25  $\mu$ V
- LOW OFFSET VOLTAGE DRIFT 0.4  $\mu$ V/ $^{\circ}$ C
- LOW NOISE 9nV/ $\sqrt{\text{Hz}}$
- OPEN LOOP GAIN 10<sup>7</sup>
- BANDWIDTH (UNITY GAIN) 2.5MHz
- ALL BIPOLAR CONSTRUCTION

### APPLICATIONS

- HIGH GAIN INSTRUMENTATION
- PRECISION DATA ACQUISITION
- PRECISION INTEGRATORS
- BIOMEDICAL AMPLIFIERS
- PRECISION THRESHOLD DETECTORS

### DESCRIPTION

HA-5130/5135 are precision operational amplifiers manufactured using a combination of key technological advancements to provide outstanding input characteristics.

A Super Beta input stage is combined with laser trimming, dielectric isolation, and matching techniques to produce 25  $\mu$ V (Max.) input offset voltage and 0.4  $\mu$ V/ $^{\circ}$ C input offset voltage average drift. Other features enhanced by this process include 9nV (Typ.) Input Noise Voltage, 1nA Input Bias Current, and 140dB Open Loop Gain.

These features coupled with 120dB CMRR and PSRR make HA-5130/5135 an ideal device for precision DC instrumentation amplifiers. Excellent input characteristics in conjunction with 2.5MHz bandwidth and 0.8V/ $\mu$ s slew rate, makes this amplifier extremely useful for precision integrator and biomedical amplifier designs. These amplifiers are also well suited for precision data acquisition and for accurate threshold detector applications.

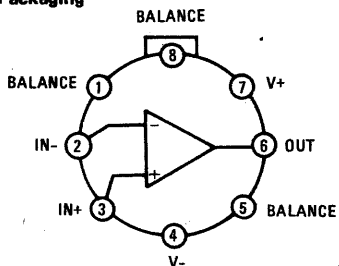
HA-5130/35 is packaged in an 8 pin (TO-99) can and an 8 lead Cerdip and is pin compatible with many existing op amp configurations.

HA-5130/5135-2 is specified for -55 $^{\circ}$ C to +125 $^{\circ}$ C operation while HA-5130/5135-5 operate from 0 $^{\circ}$ C to +75 $^{\circ}$ C.

### PINOUT

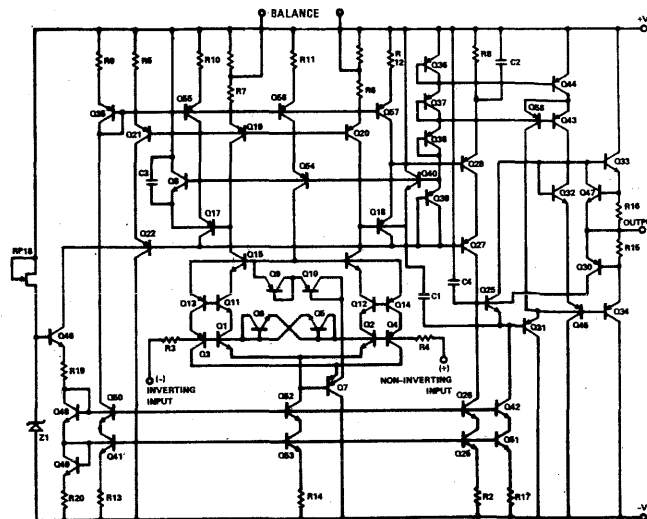
Section 11 for Packaging

TOP VIEW



\* Pins 5 and 8 are internally connected

### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

$T_A = +25^\circ\text{C}$ Unless otherwise stated		Power Dissipation (Note 2)	300mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$
Differential Input Voltage	$\pm 15.0\text{V}$	HA-5130/5135-2	$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$
		HA-5130/5135-5	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$
Output Short Circuit Duration	Indefinite	Storage Temperature Range	

## ELECTRICAL CHARACTERISTICS $V_+ = 15\text{V}, V_- = -15\text{V}$

PARAMETER	TEMP.	HA-5130-2/-5			HA-5135-2/-5			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		10	25		10	75	$\mu\text{V}$
	Full		50	60		50	130	$\mu\text{V}$
Average Offset Voltage Drift	Full		0.4	0.6		0.4	1.3	$\mu\text{V}/^\circ\text{C}$
Bias Current	+25°C		$\pm 1$	$\pm 2$		$\pm 1$	$\pm 4$	nA
	Full			$\pm 4$			$\pm 6$	nA
Bias Current Average Drift	Full		0.02	0.04		0.02	0.04	nA/°C
Offset Current	+25°C			2			4	nA
	Full			4			5.5	nA
Offset Current Average Drift	Full		0.02	0.04		0.02	0.04	nA/°C
Common Mode Range	Full	$\pm 12$			$\pm 12$			V
Differential Input Resistance	+25°C	20	30		20	30		M $\Omega$
Input Noise Voltage	+25°C			0.6			0.6	$\mu\text{V}_{\text{p-p}}$
0.1Hz to 10Hz (Note 3)								
Input Noise Voltage Density (Note 3)	+25°C							nV/ $\sqrt{\text{Hz}}$
$f_0 = 10\text{Hz}$			13.0	18.0		13.0	18.0	
$f_0 = 100\text{Hz}$			10.0	13.0		10.0	13.0	
$f_0 = 1000\text{Hz}$			9.0	11.0		9.0	11.0	
Input Noise Current (Note 3)	+25°C		15	30		15	30	pA $_{\text{p-p}}$
0.1Hz to 10Hz								
Input Noise Current Density (Note 3)	+25°C							pA/ $\sqrt{\text{Hz}}$
$f_0 = 10\text{Hz}$			0.4	0.8		0.4	0.8	
$f_0 = 100\text{Hz}$			0.17	0.23		0.17	0.23	
$f_0 = 1000\text{Hz}$			0.14	0.17		0.14	0.17	
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 4)	+25°C	120	140		120	140		dB
	Full	120			120			dB
Common Mode Rejection Ratio (Note 5)	Full	110	120		106	120		dB
Closed Loop Bandwidth ( $A_{\text{VCL}} = +1$ )	+25°C	0.6	2.5		0.6	2.5		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 6)	+25°C	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
	Full	$\pm 10$			$\pm 10$			V
Full Power Bandwidth (Note 7)	+25°C	8	10		8	10		kHz
Output Current (Note 8)	+25°C	$\pm 25$	$\pm 30$		$\pm 25$	$\pm 30$		mA
Output Resistance (Note 9)	+25°C		45			45		$\Omega$
<b>TRANSIENT RESPONSE (Note 10)</b>								
Rise Time	+25°C		340			340		ns
Slew Rate	+25°C	0.5	0.8		0.5	0.8		V/ $\mu\text{s}$
Settling Time (Note 11)	+25°C		11			11		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	Full		1.0	1.3		1.0	1.7	mA
Power Supply Rejection Ratio (Note 12)	Full	100	130		94	130		dB

### NOTES:

- Absolute maximum ratings are limiting values, applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- Derate at 6.8mW/°C for operation at ambient temp.'s above +75°C.
- Not tested. 90% of units meet or exceed these specifications.
- $V_{\text{OUT}} = \pm 10\text{V}$ ;  $R_L = 2\text{k}$ . Gain dB =  $20 \log_{10}$  Average  
 $\therefore 120\text{dB} = 1000\text{V/mV}$   
 $140\text{dB} = 10,000\text{V/mV}$
- $V_{\text{CM}} = \pm 10\text{V DC}$
- $R_L = 600\Omega$
- $R_L = 2\text{k}$ ; Full power bandwidth guaranteed based on slew rate measurement using  $\text{FPBW} = \frac{\text{SLEW RATE}}{2\pi V_{\text{PEAK}}}$
- $V_{\text{OUT}} = 10\text{V}$
- Output resistance measured under open loop conditions ( $f = 100\text{Hz}$ )
- Refer to test circuits section of the data sheet.
- Settling time is measured to 0.1% of final value for a 10V output step and  $A_V = -1$ .
- $V_{\text{SUPP}} = \pm 5\text{V DC}$  to  $\pm 20\text{V DC}$ .



## ADVANCE

### Ultra-Low Power Operational Amplifier

#### FEATURES

- LOW SUPPLY CURRENT 60  $\mu$ A
- WIDE OPERATING VOLTAGE RANGE 2V to 30V
- SINGLE SUPPLY OPERATION
- HIGH SLEW RATE 1.5V/ $\mu$ s
- HIGH GAIN 100K V/V
- AVAILABLE IN SINGLES, DUALS AND QUADS

#### APPLICATIONS

- PORTABLE INSTRUMENTS
- METER AMPLIFIERS
- TELEPHONE HEADSETS
- MICROPHONE AMPLIFIERS
- INSTRUMENTATION

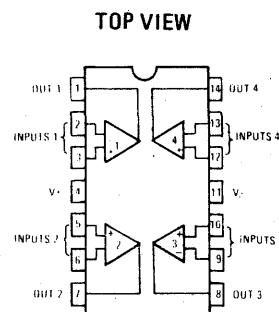
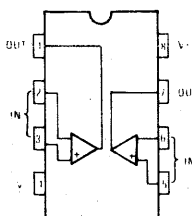
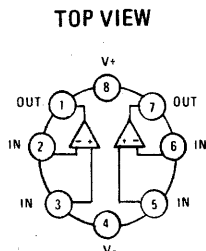
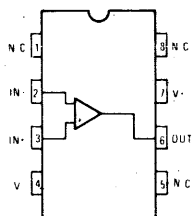
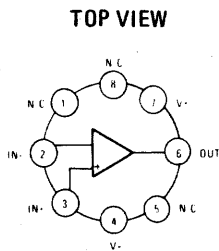
#### DESCRIPTION

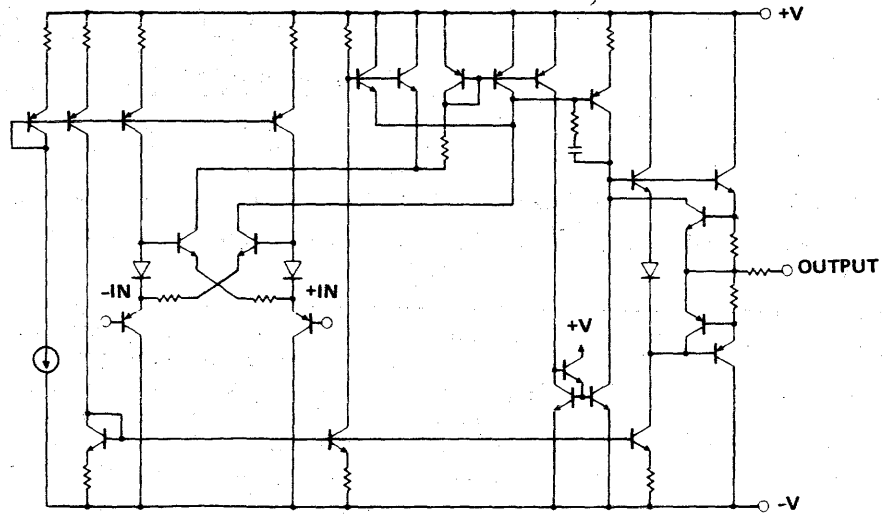
The HA-5141/42/44 ultra-low power operational amplifiers provide AC and DC performance characteristics similar to or better than most general purpose amplifiers while only drawing 1/30 of the supply current of most general purpose amplifiers. These amplifiers are well suited to applications which require low power dissipation and good electrical characteristics.

The HA-5141/42/44 provides accurate signal processing by virtue of its low input offset voltage (0.5mV), low input bias current (50nA), high open loop gain (100KV/V) and low noise, for low power operational amplifiers (20nV/ $\sqrt{\text{Hz}}$ ). These characteristics coupled with 1.5V/ $\mu$ s slew rate and 400KHz bandwidth make the HA-5141/42/44 ideal for use in low power instrumentation, audio amplifier and active filter designs. The wide range of supply voltages (2V to 30V) also allow these amplifiers to be very useful in low voltage battery powered equipment.

These amplifiers are available in singles (HA-5141, can or minidip), duals (HA-5142, can or minidip) or quads (HA-5144, 14 pin dip) with industry standard pinouts which allow the HA-5141/42/44's to be interchangeable with most other operational amplifiers.

#### PINOUTS





**SPECIFICATIONS**

**ABSOLUTE MAXIMUM RATINGS**

Voltage Between V+ and V- Terminals	40V	Operating Temperature Range	$0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$
Differential Input Voltage	$\pm 7\text{V}$		$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
Output Current	S/C Protected	Storage Temperature Range	$-65^{\circ} \leq T_A \leq +150^{\circ}\text{C}$
Internal Power Dissipation	500mW		

**ELECTRICAL CHARACTERISTICS  $V_+ = +5\text{V}$**

PARAMETER	TEMP.	HA-5141/42/44A			HA-5141/42/44			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		0.5	6		0.7	7	mV
	Full			8			9	mV
Bias Current	+25°C		45	75		45	100	nA
	Full			100			125	nA
Offset Current	+25°C		0.3	10		0.3	10	nA
	Full			15			20	nA
Common Mode Range	Full	0 to 4			0 to 3			V
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain	+25°C	50	100		20	100		KV/V
(Note 1)	Full	30			15			KV/V
Common Mode Rejection	Full	80	105		77	105		dB
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing	Full	0 to 4			0 to 3			V
(Note 1)								
<b>TRANSIENT RESPONSE</b>								
Slew Rate	+25°C	1	1.5		0.5	1		V/ $\mu\text{s}$
(Notes 1,2,3)								
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current (per Amplifier)	+25°C		45	65		50	80	$\mu\text{A}$
	Full			75			100	$\mu\text{A}$
Power Supply Rejection Ratio	Full	80	105		77	105		dB

- NOTES:  
 1.  $R_L = 50\text{K}$   
 2.  $C_L = 50\text{pf}$   
 3.  $V_{IN} = +3\text{V}$  Pulse

### FEATURES

- WIDE GAIN BANDWIDTH 100MHz
- HIGH SLEW RATE 120V/ $\mu$ s
- SETTLING TIME (0.2%) 280ns
- POWER BANDWIDTH 1000kHz
- OFFSET VOLTAGE 1.0mV
- BIAS CURRENT 20pA

### DESCRIPTION

The HA-5160/5162 is a wideband, uncompensated, operational amplifier manufactured with FET/Bipolar technologies and dielectric isolation. This monolithic amplifier features superior high frequency capabilities further enhanced by precision laser trimming of the input stage to provide excellent input characteristics. This device has excellent phase margin at a closed loop gain of 10 without external compensation.

The HA-5160/5162 offers a number of important advantages over similar FET input op amps from other manufacturers. In addition to superior bandwidth and settling characteristics, the HARRIS devices have nearly constant slew rate, bandwidth, and settling characteristics over the operating temperature range. This provides the user predictable performance in applications where settling time, full power bandwidth, closed loop bandwidth, or phase shift is critical. Note also that HARRIS specified all parameters at ambient (rather than junction) temperature to provide the designer meaningful data to predict actual operating performance.

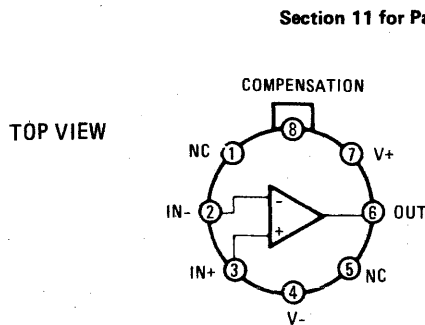
Complementing the HA-5160/5162's predictable and excellent dynamic characteristics are very low input offset voltage, very low input bias current, and extremely high input impedance. This ideal combination of features make these amplifiers most suitable for precision, high speed, data acquisition system designs and for a wide variety of signal conditioning applications.\* The HA-5160 provides excellent performance for applications which require both precision and high speed performance. The HA-5162 meets or exceeds the performance specifications of National's hybrid op amp, the LH0062.

\* -2 denotes a range of -55°C to +125°C and -5 denotes a 0°C to +75°C range.

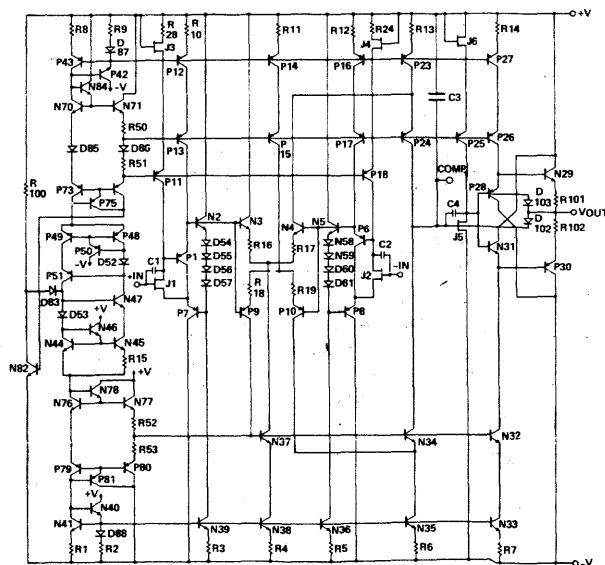
### APPLICATIONS

- VIDEO AND RF AMPLIFIERS
- DATA ACQUISITION
- PULSE AMPLIFIERS
- PRECISION SIGNAL GENERATION

### PINOUT



### SCHEMATIC



# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS

Voltage Between V+ and V-	40V
Differential Input Voltage	± 40V
Peak Output Current	Full Short Circuit Protection
Internal Power Dissipation (Note 2)	675mW
Storage Temperature Range	-65°C to +150°C

## ELECTRICAL CHARACTERISTICS V+ = +15V, V- = -15V

PARAMETER	TEMP	HA-5160-2 -55°C to +125°C			HA-5160-5 0°C to +75°C			HA-5162-5 0°C to 75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	MAX	TYP	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage	+25°C		1.0	3.0		1.0	3.0		3	15	mV
	Full		3.0	5.0		3.0	5.0		5	20	mV
Offset Voltage Average Drift	Full		10			20			20	35	μV/°C
Bias Current	+25°C		20	50		20	50		20	65	pA
	Full		5	10			10			10	nA
Offset Current	+25°C		2	10		2	10		2	10	pA
	Full		2	5		2	5		2	5	nA
Input Resistance	+25°C		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω
Common Mode Range	Full	± 10	± 11		± 10	± 11		± 10	± 11		V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain (Note 3)	+25°C	75K	150K		75K	150K		25K	100K		V/V
	Full	60K	100K		60K	100K		25K	75K		V/V
Common Mode Rejection Ratio (Note 4)	Full	74	80		74	80		70	80		dB
Gain Bandwidth Product (A <sub>v</sub> = 10)	Full		100			100			100		MHz
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Swing (Note 5)	+25°C	± 10	± 11		± 10	± 11		± 10	± 11		V
	Full	± 10	± 11		± 10	± 11		± 10	± 11		V
Output Current (Note 6)	+25°C	± 15	± 20		± 15	± 20		± 15	± 20		mA
Full Power Bandwidth (Note 7)	+25°C		1000			1000			1000		kHz
Output Resistance (Note 8)	+25°C		50			50			50		Ω
<b>TRANSIENT RESPONSE (Note 9)</b>											
Rise Time (A <sub>v</sub> = 10)	+25°C		20			20			20		ns
Slew Rate (A <sub>v</sub> = 10)	+25°C	100	120		100	120		50	70		V/μs
Settling Time (Note 10)	+25°C		280			280			400		ns
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current	Full		8.0	10		8.0	10		8.0	12	mA
Power Supply Rejection Ratio (Note 11)	+25°C	74	86		74	86		70	86		dB

LINEAR

Harris Semiconductor

## Precision JFET Input Operational Amplifier

### FEATURES

- LOW OFFSET VOLTAGE 100  $\mu$ V
- LOW OFFSET VOLTAGE DRIFT 2  $\mu$ V/ $^{\circ}$ C
- LOW NOISE 10nV/ $\sqrt{\text{Hz}}$
- HIGH OPEN LOOP GAIN 600K V/V
- WIDE BANDWIDTH 5MHz

### APPLICATIONS

- HIGH GAIN INSTRUMENTATION AMPLIFIERS
- PRECISION DATA ACQUISITION
- PRECISION INTEGRATORS
- PRECISION THRESHOLD DETECTORS

### DESCRIPTION

The Harris HA-5170 is a precision, JFET input, operational amplifier which features low noise, low offset voltage and low offset voltage drift. Constructed using FET/Bipolar technology, the Harris Dielectric Isolation (DI) process, and laser trimming this amplifier offers low input bias and offset currents. This operational amplifier design also completely eliminates the troublesome errors due to warm-up drift.

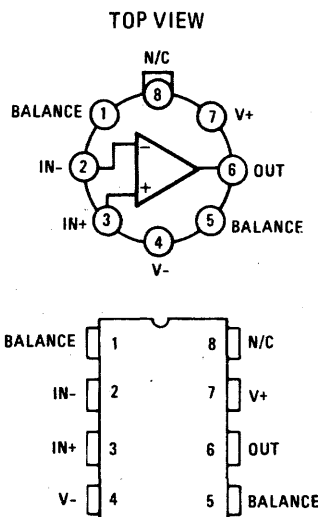
Complementing these excellent input characteristics are dynamic performance characteristics never before available from precision operational amplifiers. An 8V/ $\mu$ s slew rate and 5MHz bandwidth allow the designer to extend precision instrumentation applications in both speed and bandwidth. These characteristics make the HA-5170 well suited for precision integrator amplifier designs.

The superior input characteristics also make the HA-5170 ideally suited for transducer signal amplifiers, precision voltage followers and precision data acquisition systems.

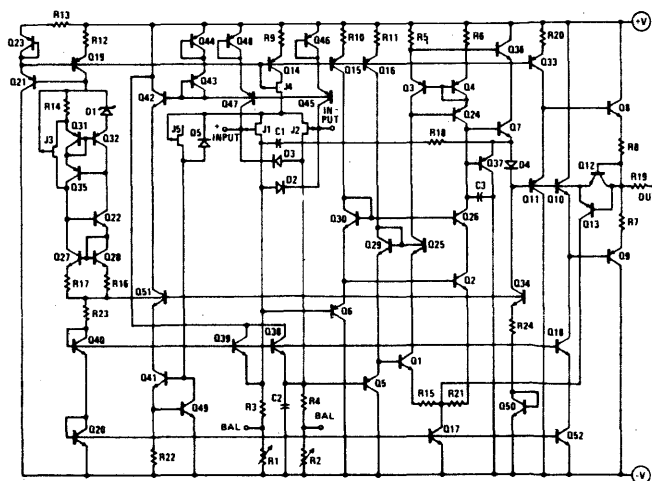
Packaged in an 8-pin (TO-99) can or an 8 lead Minidip, the HA-5170 is pin compatible with most existing op amp configurations.

### PINOUT

Section 11 for Packaging



### SCHEMATIC



Harris Semiconductor LINEAR

# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

$T_A = +25^\circ\text{C}$ Unless otherwise stated		Power Dissipation (Note 2)	675mW
Voltage Between V+ and V-Terminals	44.0V	Operating Temperature Range	
Differential Input Voltage	$\pm 30.0\text{V}$	HA-5170-2	$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$
		HA-5170-5	$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$
Output Short Circuit Duration	Indefinite	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

## ELECTRICAL CHARACTERISTICS $V_+ = 15\text{V}, V_- = -15\text{V}$

PARAMETER	TEMP.	HA-5170-2			HA-5170-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		0.1	0.3		0.1	0.3	mV
	Full			0.5			0.5	mV
Average Offset Voltage Drift (Note 3)	Full		2	5		2	5	$\mu\text{V}/^\circ\text{C}$
Bias Current	+25°C		20	100		20	100	pA
	Full		3	30		0.1	2	nA
Bias Current Average Drift	Full		3			3		$\text{pA}/^\circ\text{C}$
Offset Current	+25°C		3	30		3	60	pA
	Full			5			0.1	nA
Offset Current Average Drift	Full		0.3			0.3		$\text{pA}/^\circ\text{C}$
Common Mode Range	Full	$\pm 10$	+15.1 -12		$\pm 10$	+15.1 -12		V
Differential Input Resistance	+25°C		$6 \times 10^{10}$			$6 \times 10^{10}$		$\Omega$
Input Capacitance	+25°C		12			12		pF
Input Noise Voltage 0.1Hz to 10Hz (Note 3)	+25°C		0.5	5		0.5	5	$\mu\text{V}_{\text{p-p}}$
Input Noise Voltage Density (Note 3)	+25°C							$\text{nV}/\sqrt{\text{Hz}}$
$f_0 = 10\text{Hz}$			20	150		20	150	$\text{nV}/\sqrt{\text{Hz}}$
$f_0 = 100\text{Hz}$			12	50		12	50	$\text{nV}/\sqrt{\text{Hz}}$
$f_0 = 1000\text{Hz}$			10	25		10	25	$\text{nV}/\sqrt{\text{Hz}}$
Input Noise Current Density (Note 3)	+25°C							$\text{pA}/\sqrt{\text{Hz}}$
$f_0 = 10\text{Hz}$			.05			.05		$\text{pA}/\sqrt{\text{Hz}}$
$f_0 = 100\text{Hz}$			.01			.01		$\text{pA}/\sqrt{\text{Hz}}$
$f_0 = 1000\text{Hz}$			.01			.01		$\text{pA}/\sqrt{\text{Hz}}$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 4)	+25°C		300K	600K		300K	600K	V/V
	Full		200K			250K		V/V
Common Mode Rejection Ratio (Note 5)	Full	85	100		90	100		dB
Closed Loop Bandwidth ( $A_{\text{VCL}} = +1$ )	+25°C	4	8		4	8		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 6)	+25°C	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
Full Power Bandwidth (Note 7)	+25°C	80	120		80	120		kHz
Output Current (Note 8)	+25°C	$\pm 10$			$\pm 10$			mA
Output Resistance (Note 9)	+25°C		45			45		$\Omega$
<b>TRANSIENT RESPONSE</b>								
Rise Time	+25°C		45	100		45	100	ns
Slew Rate	+25°C	5	8		5	8		$\text{V}/\mu\text{s}$
Settling Time (Note 10)	+25°C		1			1		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	Full		1.9	2.5		1.9	2.5	mA
Power Supply Rejection Ratio (Note 11)	Full	85	105		90	105		dB

### NOTES:

- Absolute maximum ratings are limiting values, applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- Derate at 6.8 mW/°C for operation at ambient temperatures above +75°C.
- Parameter is not 100% tested. 90% of all units meet or exceed these specifications.
- $V_{\text{OUT}} = \pm 10\text{V}; R_L = 2\text{k}\Omega$ .
- $V_{\text{CM}} = \pm 10\text{V D. C.}$
- $R_L = 2\text{k}\Omega$ .
- $R_L = 2\text{k}\Omega$ ; Full power bandwidth guaranteed based on  
slew rate measurement using  $\text{FPBW} = \frac{\text{SLEW RATE}}{2.77V_{\text{PEAK}}}$
- $V_{\text{OUT}} = 10\text{V}$ .
- Output resistance measured under open loop conditions ( $f = 100\text{Hz}$ ).
- Settling time is measured to 0.1% of final value for a 10V output step and  $A_V = -1$ .
- $V_{\text{SUPP}} = \pm 5\text{V D. C. to } \pm 20\text{V D. C.}$

# HA-5180/5180A

*Low Bias Current, Low Power  
JFET Input Operational Amplifier*

APRIL 1982

**Preliminary**
**FEATURES**

- ULTRA LOW BIAS CURRENT                      250fA
- LOW POWER SUPPLY CURRENT                0.8mA
- LOW OFFSET VOLTAGE                        0.5mV (max.)
- BANDWIDTH                                      2 MHz
- SLEW RATE                                      7V/ $\mu$ s

**DESCRIPTION**

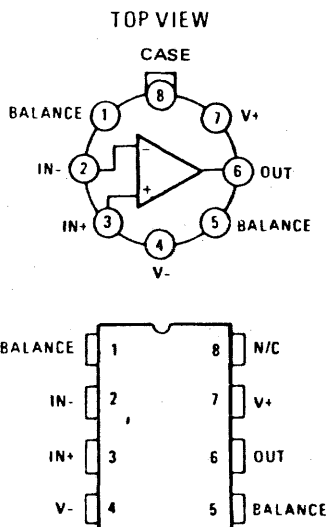
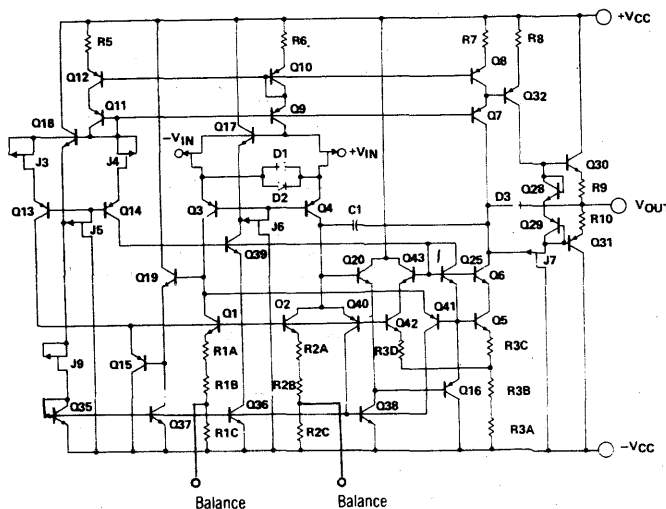
The HARRIS HA-5180/5180A is an ultra low input bias current, JFET input, monolithic operational amplifier which also features low power, low offset voltage and excellent AC characteristics. Employing FET/Bipolar construction coupled with dielectric isolation this operational amplifier offers the lowest input bias currents (250fA typ.) available in any monolithic operational amplifier. The HA-5180/5180A has another unique feature in which the offset bias current may be nulled by externally adjusting the offset voltage. For applications which require precision performance the HA-5180A offers an input offset voltage of 0.5 mV (max) while the HA-5180 offers 3 mV (max.)

The HA-5180/5180A also offers excellent AC performance not previously available in similar hybrid or monolithic op amp designs. The 2 MHz bandwidth and 7V/ $\mu$ s slew rate of the HA-5180/5180A extends the bandwidth and speed for applications such as very low drift sample and hold amplifiers and photo-current detectors. Other applications include use in electrometer designs, pH/Ion sensitive electrodes, low current oxygen sensors, long term precision integrators and very high impedance buffer measurement designs.

The HA-5180/5180A is packaged in an 8-pin (TO-99) can and an 8-lead cerdip and is pin compatible with most existing op amp configurations. The case of the TO-99 package is internally connected to pin 8 so that it may be connected to the same potential as the input. This feature helps minimize stray leakage to the case, helps shield the amplifier from external noise and reduces common mode input capacitance.

**APPLICATIONS**

- ELECTROMETER AMPLIFIER DESIGNS
- PHOTO CURRENT DETECTORS
- PRECISION LONG-TERM INTEGRATORS
- LOW DRIFT SAMPLE & HOLD CIRCUITS
- VERY HIGH IMPEDANCE BUFFERS
- HIGH IMPEDANCE BIOLOGICAL MICRO PROBES

**PINOUT**

**SCHEMATIC**


# SPECIFICATIONS



## ABSOLUTE MAXIMUM RATINGS (Note 1)

$T_A = +25^\circ\text{C}$ Unless otherwise stated		Power Dissipation (Note 2)	300mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	
Differential Input Voltage	$\pm 40\text{V}$	HA-5180/5180A-2	$-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$
		HA-5180/5180A-5	$0^\circ\text{C} \leq T_A \leq 75^\circ\text{C}$
Output Short Circuit Duration	Indefinite	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

## ELECTRICAL RATINGS

V+ = 15V, V- = -15V

PARAMETER	TEMP.	5180A-2			5180A-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		0.1	0.5		0.1	0.5	mV
	Full			1			1	mV
Average Offset Voltage Drift	Full		5			5		$\mu\text{V}/^\circ\text{C}$
Bias Current (Note 3)	+25°C		250	1000		250	1000	fA
	Full		100	500		6	30	pA
Offset Current (Note 3)	+25°C		30	200		30	200	fA
	Full		6	30		1	5	pA
Common Mode Range	Full	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
Differential Input Resistance	+25°C		$10^{12}$			$10^{12}$		$\Omega$
Input Noise Voltage (f=1kHz)	+25°C		70			70		$\text{nV}/\sqrt{\text{Hz}}$
Input Noise Current (f=1kHz)	+25°C		0.01			0.01		$\text{pA}/\sqrt{\text{Hz}}$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 4)	+25°C	200k	1M		200k	1M		V/V
	Full	150k			150k			V/V
Common Mode Rejection Ratio (Note 5)	Full	90	110		90	110		dB
Closed Loop Bandwidth ( $A_{VCL} = +1$ )	+25°C		2			2		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 6)	+25°C	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
	Full	$\pm 10$			$\pm 10$			V
Full Power Bandwidth (Note 7)	+25°C		110			110		kHz
Output Current (Note 8)	+25°C	$\pm 10$	$\pm 15$		$\pm 10$	$\pm 15$		mA
Output Resistance (Note 9)	+25°C		25			25		$\Omega$
<b>TRANSIENT RESPONSE</b>								
Rise Time	+25°C		75			75		ns
Slew Rate	+25°C	4	7		4	7		V/ $\mu\text{s}$
Settling Time (Note 10)	+25°C		2			2		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	Full		0.7	1		0.8	1	mA
Power Supply Rejection Ratio (Note 11)	Full	85	105		85	105		dB

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ELECTRICAL RATINGS

$V_+ = 15V, V_- = -15V$

PARAMETER	TEMP	5180A-2			5180A-5			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		1	3		1	3	mV
	Full			4			4	mV
								mV
								mV
Average Offset Voltage Drift	Full		5			5		$\mu V/^\circ C$
Bias Current (Note 3)	+25°C		250	1000		250	1000	fA
	Full		100	500		6	30	pA
Offset Current (Note 3)	+25°C		30	200		30	200	fA
	Full		6	30		1	5	pA
Common Mode Range	Full	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
Differential Input Resistance	+25°C		$10^{12}$			$10^{12}$		$\Omega$
Input Noise Voltage (f = 1kHz)	+25°C		70			70		$nV/\sqrt{Hz}$
Input Noise Current (f = 1kHz)	+25°C		0.01			0.01		$pA/\sqrt{Hz}$
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 4)	+25°C	200k	1M		200k	1M		V/V
	Full	150k			150k			V/V
Common Mode Rejection Ratio (Note 5)	Full	90	110		90	110		dB
Closed Loop Bandwidth ( $A_{VCL} = +1$ )	+25°C		2			2		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 6)	+25°C	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
	Full	$\pm 10$			$\pm 10$			V
Full Power Bandwidth (Note 7)	+25°C		110			110		kHz
Output Current (Note 8)	+25°C	$\pm 10$	$\pm 15$		$\pm 10$	$\pm 15$		mA
Output Resistance (Note 9)	+25°C		25			25		$\Omega$
<b>TRANSIENT RESPONSE</b>								
Rise Time	+25°C		75			75		ns
Slew Rate	+25°C	4	7		4	7		V/ $\mu s$
Settling Time (Note 10)	+25°C		2			2		$\mu s$
<b>POWER SUPPLY CHARACTERISTICS</b>								
Supply Current	Full		0.7	1		0.8	1	mA
Power Supply Rejection Ratio (Note 11)	Full	85	105		85	105		dB

NOTES:

- Absolute maximum ratings are limiting values, applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- Derate at 6.9 mW/°C for operation at ambient temperatures above +75°C.
- This parameter is guaranteed by design and is not 100% tested.
- $V_{OUT} = \pm 10V$ ;  $R_L = 2k$ . Gain dB =  $20 \log_{10} A_v$ .
- $V_{CM} = \pm 10V$  D.C.
- $R_I = 2k$
- $R_L = 2k$ ; Full power bandwidth guaranteed based on slew rate measurement using  $FPBW = \frac{SLEW\ RATE}{2\pi V_{PEAK}}$
- $V_{OUT} = \pm 10V$ .
- Output resistance measured under open loop conditions (f = 100Hz)
- Settling time is measured to 0.1% of final value for a 10V output step and  $A_v = -1$ .
- $V_{SUPP} = +5V$  D.C. to  $\pm 20V$  D.C.

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## Wideband, Fast Settling Operational Amplifier

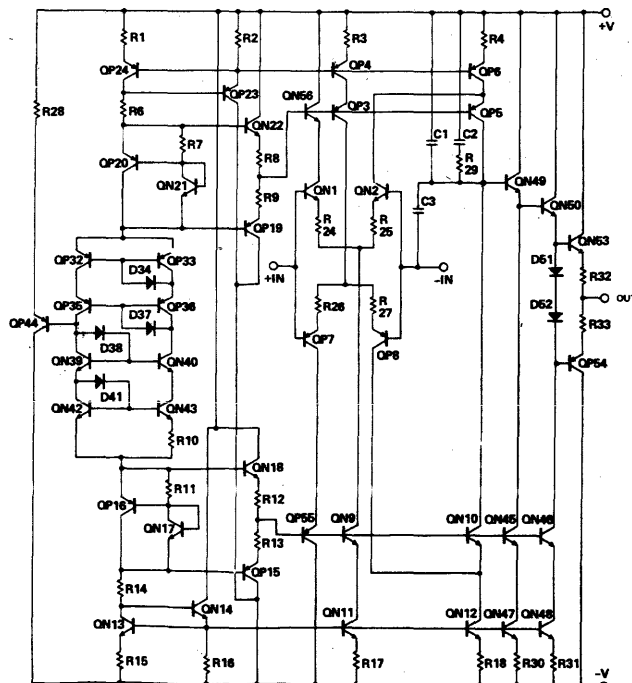
### FEATURES

- FAST SETTLING TIME 70ns
- VERY HIGH SLEW RATE 200V/ $\mu$ s
- WIDE GAIN-BANDWIDTH 150MHz
- POWER BANDWIDTH 6.5MHz
- LOW OFFSET VOLTAGE 5mV
- INPUT VOLTAGE NOISE 15nV/ $\sqrt{\text{Hz}}$
- MONOLITHIC BIPOLAR CONSTRUCTION

### APPLICATIONS

- FAST, PRECISE D/A CONVERTERS
- HIGH SPEED SAMPLE-HOLD CIRCUITS
- PULSE AND VIDEO AMPLIFIERS
- WIDEBAND AMPLIFIERS
- REPLACE COSTLY HYBRIDS

### SCHEMATIC



### GENERAL DESCRIPTION

HA-5190/5195 are monolithic operational amplifiers featuring an ultimate combination of speed, precision, and bandwidth. Employing monolithic bipolar construction coupled with dielectric isolation, these devices are capable of delivering an unparallelled 200V/ $\mu$ s slew rate with a settling time of 70ns (0.1%, 5V output step). These truly differential amplifiers are designed to operate at gains  $\geq 5$  without the need for external compensation. Other outstanding HA-5190/5195 features are 150MHz gain-bandwidth-product and 6.5MHz full power bandwidth. In addition to these dynamic characteristics, these amplifiers also have excellent input characteristics such as 5mV offset voltage and 15nV input voltage noise (at 1kHz).

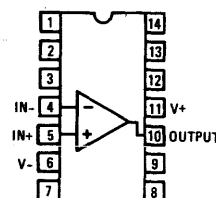
With 200V/ $\mu$ s slew rate and 70ns settling time, these devices make ideal output amplifiers for accurate, high speed D/A converters or the main components in high speed sample/hold circuits. 150MHz gain-bandwidth-product, 6.5MHz power bandwidth, and 5mV offset voltage make HA-5190/5195 ideally suited for a variety of pulse and wideband video amplifier applications.

At temperatures above +75°C, a heat sink is required for HA-5190. (See note 2). HA-5190 is specified over the -55°C to +125°C range while HA-5195 is specified from 0°C to +75°C.

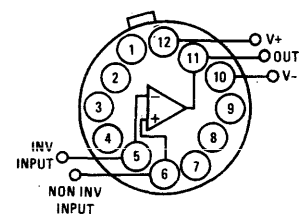
### PINOUTS

Section 11 for Packaging

TOP VIEW



TOP VIEW



CASE TIED TO V-  
LH0032 PINOUT



**ABSOLUTE MAXIMUM RATINGS (Note 1)**

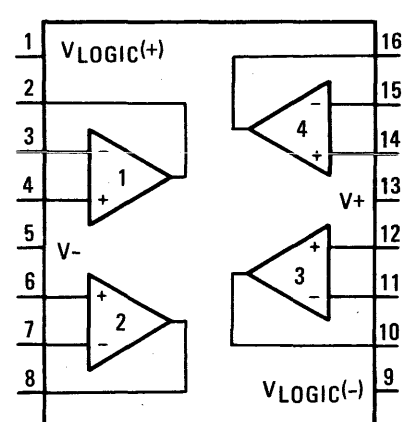
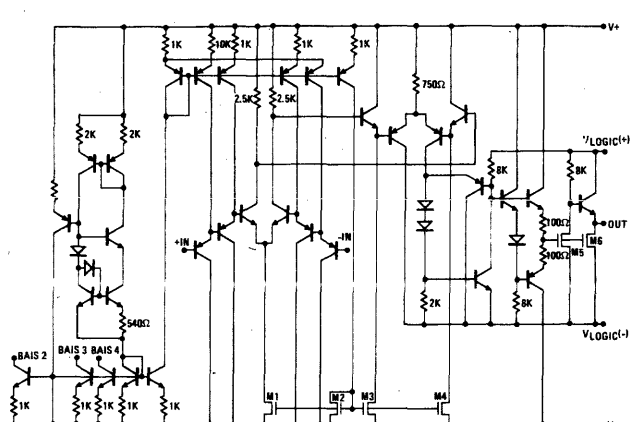
Voltage between V+ and V- Terminals	35V
Differential Input Voltage	6V
Output Current	50mA (Peak)
Internal Power Dissipation (Note 2)	870mW (Cerdip); 1W (TO-8) Free Air
Operating Temperature Range: (HA-5190)	$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
(HA-5195)	$0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$
Storage Temperature Range	$-65^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$

**ELECTRICAL CHARACTERISTICS**  $V_{\text{SUPPLY}} = \pm 15$  Volts;  $R_L = 1\text{K}$  ohms, unless otherwise specified.

PARAMETER	TEMP	HA-5190 -55°C to +125°C			HA-5195 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>								
Offset Voltage	+25°C		3.0	5.0		3.0	6	mV
	FULL			10.0			10.0	mV
Average Offset Voltage Drift	FULL		20			20		μV/°C
Bias Current	+25°C		5	15		5	15	μA
	FULL			20			20	μA
Offset Current	+25°C		1	4		1	4	μA
	FULL			6			6	μA
Input Resistance	+25°C		10			10		Kohms
Input Capacitance	+25°C		1.0			1.0		pF
Common Mode Range	FULL	±5			±5			V
Input Noise Voltage (f = 1kHz, R <sub>g</sub> = 0Ω)	+25°C		15			15		nV/√Hz
<b>TRANSFER CHARACTERISTICS</b>								
Large Signal Voltage Gain (Note 3)	+25°C	15K	30K		10K	30K		V/V
	FULL	5K			5K			V/V
Common-Mode Rejection Ratio (Note 4)	FULL	74			74			dB
Gain-Bandwidth-Product (Notes 5 & 6)	+25°C		150			150		MHz
<b>OUTPUT CHARACTERISTICS</b>								
Output Voltage Swing (Note 3)	FULL	±5	±8		±5	±8		V
Output Current (Note 3)	+25°C	25	30		25	30		mA
Output Resistance	+25°C		30			30		Ohms
Full Power Bandwidth (Note 3 & 7)	+25°C	5	6.5		5	6.5		MHz
<b>TRANSIENT RESPONSE (Note 8)</b>								
Rise Time	+25°C		13	18		13	18	ns
Overshoot	+25°C		8			8		%
Slew Rate	+25°C	160	200		160	200		V/μs
Settling Time:								
5V Step to 0.1%	+25°C		70			70		ns
5V Step to 0.01%	+25°C		100			100		ns
2.5V Step to 0.1%	+25°C		50			50		ns
2.5V Step to 0.01%	+25°C		80			80		ns
<b>POWER REQUIREMENTS</b>								
Supply Current	FULL		19	28		19	28	mA
Power Supply Rejection Ratio (Note 9)	FULL	70	90		70	90		dB

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FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>FAST RESPONSE TIME 130ns</li> <li>LOW OFFSET VOLTAGE 2.0mV</li> <li>LOW OFFSET CURRENT 10nA</li> <li>SINGLE OR DUAL-VOLTAGE SUPPLY OPERATION</li> <li>SELECTABLE OUTPUT LOGIC LEVELS</li> <li>ACTIVE PULL-UP/PULL-DOWN OUTPUT CIRCUIT - NO EXTERNAL RESISTORS REQUIRED</li> </ul>	<p>The HA-4900 series are monolithic, quad, precision comparators offering fast response time, low offset voltage, low offset current, and virtually no channel-to-channel crosstalk for applications requiring accurate, high speed, signal level detection. These comparators can sense signals at ground level while being operated from either a single +5 volt supply (digital systems) or from dual supplies (analog networks) up to <math>\pm 15</math> volts. The HA-4900 series contains a unique current driven output stage which can be connected to logic system supplies (<math>V_{Logic+}</math> and <math>V_{Logic-}</math>) to make the output levels directly compatible (no external components needed) with any standard logic or special system logic levels. In combination analog/digital systems, the design employed in the HA-4900 series input and output stages prevents troublesome ground coupling of signals between analog and digital portions of the system.</p> <p>These comparators' combination of features makes them ideal components for signal detection and processing in data acquisition systems, test equipment, and microprocessor/analog signal interface networks.</p> <p>All devices are available in 16 pin dual-in-line ceramic packages. The HA-4900/4902-2 operates from <math>-55^{\circ}\text{C}</math> to <math>+125^{\circ}\text{C}</math> and the HA-4905-5 operates over a <math>0^{\circ}\text{C}</math> to <math>+75^{\circ}\text{C}</math> temperature range.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>THRESHOLD DETECTOR</li> <li>ZERO-CROSSING DETECTOR</li> <li>WINDOW DETECTOR</li> <li>ANALOG INTERFACES FOR MICROPROCESSORS</li> <li>HIGH STABILITY OSCILLATORS</li> <li>LOGIC SYSTEM INTERFACES</li> </ul>	
PINOUT	SCHEMATIC
<p style="text-align: center;">Section 11 for Packaging</p> <p style="text-align: center;">Top View</p> 	 <p style="text-align: center;">ONE FOURTH ONLY (HA-4900 SERIES)</p>

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**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Voltage Between V+ and V-	33V
Voltage Between V <sub>Logic(+)</sub> and V <sub>Logic(-)</sub>	18V
Differential Input Voltage	±15V
Peak Output Current	±50mA
Internal Power Dissipation (Note 7, 8)	880mW
Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ 150°C

**ELECTRICAL CHARACTERISTICS** V+ = +15.0V, V- = -15.0V, V<sub>Logic(+)</sub> = 5.0V, V<sub>Logic(-)</sub> = GND.

PARAMETER	TEMP	HA-4900-2 -55°C to +125°C			HA-4902-2 -55°C to +125°C			HA-4905-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>											
Offset Voltage (Note 2)	25°C		2	3		2	5		4	7.5	mV
	Full			4			8			10	mV
Offset Current	25°C		10	25		10	35		25	50	nA
	Full			35			35			70	nA
Bias Current (Note 3)	25°C		50	75		50	150		100	150	nA
	Full			150			200			300	nA
Input Sensitivity (Note 4)	25°C			V <sub>io</sub> +.3			V <sub>io</sub> +.5			V <sub>io</sub> +.5	mV
	Full			V <sub>io</sub> +.4			V <sub>io</sub> +.6			V <sub>io</sub> +.7	mV
Common Mode Range	Full	V-		V+ -2.4	V-		V+ +2.4	V-		V+ +2.4	V
<b>TRANSFER CHARACTERISTICS</b>											
Large Signal Voltage Gain	25°C		400K			400K			400K		V/V
Response Time (T <sub>pd0</sub> ) (Note 5)	25°C		130	200		130	200		130	200	ns
Response Time (T <sub>pd1</sub> ) (Note 5)	25°C		180	215		180	215		180	215	ns
<b>OUTPUT CHARACTERISTICS</b>											
Output Voltage Level											
Logic "Low State" (V <sub>OL</sub> ) (Note 6)	Full		0.2	0.4		0.2	0.4		0.2	0.4	V
Logic "High State" (V <sub>OH</sub> ) (Note 6)	Full	3.5	4.2		3.5	4.2		3.5	4.2		V
Output Current											
I <sub>Sink</sub>	Full	3.0			3.0			3.0			mA
I <sub>Source</sub>	Full	3.0			3.0			3.0			mA
<b>POWER SUPPLY CHARACTERISTICS</b>											
Supply Current, I <sub>ps</sub> (+)	25°C		6.5	20		6.5	20		7	20	mA
Supply Current, I <sub>ps</sub> (-)	25°C		4	8		4	8		5	8	mA
Supply Current, I <sub>ps</sub> (Logic)	25°C		2.0	4		2.0	4		2.0	4	mA
Supply Voltage Range											
V <sub>Logic</sub> (+) (Note 7)	Full	0		+15.0	0		+15.0	0		+15.0	V
V <sub>Logic</sub> (-) (Note 7)	Full	-15.0		0	-15.0		0	-15.0		0	V

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# HARRIS

# HA-5102/04/12/14

## Low Noise High Performance

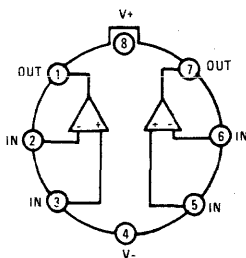
### Preliminary

### Operational Amplifiers

FEATURES	DESCRIPTION								
<ul style="list-style-type: none"> <li>● LOW NOISE <span style="float: right;">4.3 nV/√Hz</span></li> <li>● WIDE BANDWIDTH <span style="float: right;">8MHz (COMP.) 60MHz (UNCOMP.)</span></li> <li>● HIGH SLEW RATE <span style="float: right;">3V/μs (COMP.) 20V/μs (UNCOMP.)</span></li> <li>● LOW OFFSET VOLTAGE <span style="float: right;">2mV</span></li> <li>● SINGLE SUPPLY OPERATION</li> <li>● AVAILABLE IN DUALS OR QUADS</li> </ul>	<p>Low noise and high performance are key words describing HA-5102/04/12/14. These general purpose amplifiers offer an array of dynamic specifications ranging from 3V/μs slew rate and 8MHz bandwidth (5102/04) to 20V/μs slew rate and 60MHz gain-bandwidth-product (HA-5112/14). Complementing these outstanding parameters is a very low noise specification of 4.3 nV/√Hz at 1KHz.</p> <p>Fabricated using the Harris standard high frequency process, these operational amplifiers also offer excellent input specifications such as 2.0mV offset voltage and 30nA offset current. Complementing these specifications are 108dB open loop gain and 108dB channel separation. Consuming a very modest amount of power (90mW/package for duals and 150mW/package for quads), HA-5102/04/12/14 also provide the flexibility of operating from a single +5V supply.</p> <p>This impressive combination of features make this series of amplifiers ideally suited for designs ranging from audio amplifiers and active filters to the most demanding signal conditioning and instrumentation circuits.</p> <p>These operational amplifiers are available in dual or quad form with industry standard pinouts allowing for immediate interchangeability with most other dual and quad operational amplifiers.</p> <table border="0" style="width: 100%;"> <tr> <td style="width: 50%;">HA-5102</td> <td>Dual, Compensated</td> </tr> <tr> <td>HA-5112</td> <td>Dual, Uncompensated</td> </tr> <tr> <td>HA-5104</td> <td>Quad, Compensated</td> </tr> <tr> <td>HA-5114</td> <td>Quad, Uncompensated</td> </tr> </table>	HA-5102	Dual, Compensated	HA-5112	Dual, Uncompensated	HA-5104	Quad, Compensated	HA-5114	Quad, Uncompensated
HA-5102	Dual, Compensated								
HA-5112	Dual, Uncompensated								
HA-5104	Quad, Compensated								
HA-5114	Quad, Uncompensated								
APPLICATIONS									
<ul style="list-style-type: none"> <li>● HIGH Q, ACTIVE FILTERS</li> <li>● AUDIO AMPLIFIERS</li> <li>● INSTRUMENTATION AMPLIFIERS</li> <li>● INTEGRATORS</li> <li>● SIGNAL GENERATORS</li> </ul>									

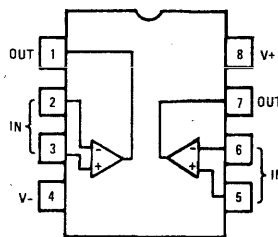
### PINOUTS

TOP VIEW



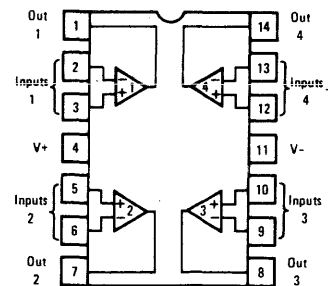
HA-5102/12

TOP VIEW



HA-5102/12

TOP VIEW



HA-5104/14

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**ABSOLUTE MAXIMUM RATINGS (Note 1)**

$T_A = +25^\circ\text{C}$ Unless Otherwise Stated		Power Dissipation (Note 4)	880mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	
Differential Input Voltage	$\pm 7\text{V}$	HA-5102/5104/5112/5114-2	$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$
Input Voltage (Note 2)	$\pm 15.0\text{V}$	HA-5102/5104/5112/5114-5	$0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}$
Output Short Circuit Duration (Note 3)	Indefinite	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$

**ELECTRICAL CHARACTERISTICS**

V+ = 15VDC; V- = -15VDC

PARAMETER	TEMP	HA-5102-2 HA-5112-2 -55°C to +125°C			HA-5104-2 HA-5114-2 -55°C to +125°C			HA-5102-5 HA-5112-5 0°C to +75°C			HA-5104-5 HA-5114-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>														
Offset Voltage	+25°C		0.5	2.0		0.5	2.5		0.5	2.0		0.5	2.5	mV
	Full			2.5			3.0			2.5			3.0	mV
Offset Voltage Average Drift	Full		3			3			3			3		$\mu\text{V}/^\circ\text{C}$
Bias Current	+25°C		130	200		130	200		130	200		130	200	nA
	Full			325			325			325			325	nA
Offset Current	+25°C		30	75		30	75		30	75		30	75	nA
	Full			125			125			125			125	nA
Input Resistance	+25°C		500			500			500			500		k $\Omega$
Common Mode Range	Full	$\pm 12$			$\pm 12$			$\pm 12$			$\pm 12$			V
<b>TRANSFER CHARACTERISTICS</b>														
Large Signal Voltage Gain (Note 5)	+25°C													V/V
	Full	100K	250K		100K	250K		100K	250K		100K	250K		
Common Mode Rejection Ratio (Note 6)	Full	86			86			86			86			dB
Small Signal Bandwidth	+25°C								8			8		MHz
Gain Bandwidth Product	+25°C									60		60		MHz
Channel Separation (Note 7)	+25°C								180			108		dB
<b>OUTPUT CHARACTERISTICS</b>														
Output Voltage Swing ( $R_L = 10\text{K}$ )	Full	$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$		V
	Full	$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		$\pm 10$	$\pm 12$		V
Output Current (Note 8)	Full	$\pm 10$	$\pm 15$		$\pm 10$	$\pm 15$		$\pm 10$	$\pm 15$		$\pm 10$	$\pm 15$		mA
Full Power Bandwidth														
(Note 9) HA-5102/5104	+25°C		50			50			50			50		kHz
HA-5112/5114	+25°C		250			250			250			250		kHz
Output Resistance	+25°C		110			110			110			110		$\Omega$

LINEAR

Harris Semiconductor

**SPECIFICATIONS**



**ELECTRICAL CHARACTERISTICS**

V+ = 15VDC; V- = -15VDC

PARAMETER	TEMP	HA-5102-2 HA-5112-2 -55°C to +125°C			HA-5104-2 HA-5114-2 -55°C to +125°C			HA-5102-5 HA-5112-5 0°C to +75°C			HA-5104-5 HA-5114-5 0°C to +75°C			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
<b>TRANSIENT RESPONSE (Note 10)</b>														
Rise time														
HA-5102/5104	+25°C		50	100		50	100		50	100		50	100	ns
HA-5112/5114	+25°C		38	60		38	60		38	60		38	60	ns
Overshoot														
HA-5102/5104	+25°C		20	35		20	35		20	35		20	35	%
HA-5112/5114	+25°C		30	40		30	40		30	40		30	40	%
Slew Rate														
HA-5102/5104	+25°C	±1	±3		±1	±3		±1	±3		±1	±3		V/μs
HA-5112/5114	+25°C	±12	±20		±12	±20		±12	±20		±12	±20		V/μs
Settling Time (Note 11)														
HA-5102/5104	+25°C		4.5			4.5			4.5			4.5		μs
HA-5112/5114	+25°C		0.6			0.6			0.6			0.6		μs
<b>NOISE CHARACTERISTICS</b>														
Input Noise Voltage	+25°C													
f = 10Hz			17			17			17			17		nV/√Hz
f = 1KHz			4.3			4.3			4.3			4.3		nV/√Hz
Input Noise Current	+25°C													
f = 10Hz			5.1			5.1			5.1			5.1		pA/√Hz
f = 1KHz			.57			.57			.57			.57		pA/√Hz
Broadband Noise Voltage	+25°C													
f = DC to -30KHz			870			870			870			870		nVrms
<b>POWER SUPPLY CHARACTERISTICS</b>														
Supply Current														
HA-5102/5112	+25°C		3.0	5.0		3.0	5.0		3.0	5.0		3.0	5.0	mA
HA-5104/5114	+25°C		5.0	6.5		5.0	6.5		5.0	6.5		5.0	6.5	mA
Power Supply Rejection Ratio (Note 6)	Full	86			86			86			86			dB

LINEAR

Harris Semiconductor



### FEATURES

- DIFFERENTIAL PHASE ERROR .1°
- DIFFERENTIAL GAIN ERROR .1%
- HIGH SLEW RATE 1000V/μs
- WIDE POWER BANDWIDTH DC TO 80MHz
- FAST RISE TIME 3ns
- WIDE POWER SUPPLY RANGE ±5 to ±16V

### DESCRIPTION

The HA-5033 is a high performance voltage follower designed for use in high frequency applications requiring a high speed wideband buffer. Specifically designed for driving coaxial cable and featuring a bandwidth of 250MHz and outstanding differential gain and phase characteristics, the HA-5033 is an excellent device for the line driver and buffer applications common to video circuit design. In addition, the HA-5033 offers a minimum slew rate of 1000V/μs.

This device is fabricated using the HARRIS dielectric isolation process. Reduced parasitic capacitance and leakage currents along with improved high frequency PNP transistor performance are among the benefits derived from the superior electrical isolation of the D.I. process.

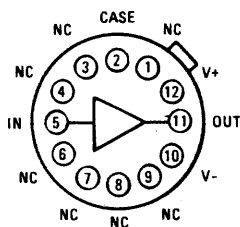
Available in a 12 pin (TO-8) metal can or an 8 pin epoxy mini-dip, the HA-5033-2 is specified over the military temperature range of -55°C to +125°C. The HA-5033-5 is characterized over the commercial temperature range of 0°C to +75°C.

### APPLICATIONS

- VIDEO BUFFER
- HIGH FREQUENCY BUFFER
- OP AMP ISOLATION BUFFER
- HIGH SPEED LINE DRIVER
- IMPEDANCE MATCHING

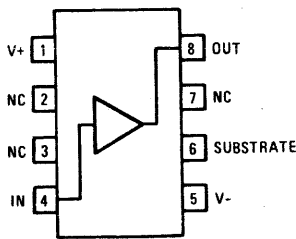
### PINOUT

#### VIDEO BUFFER PINOUT



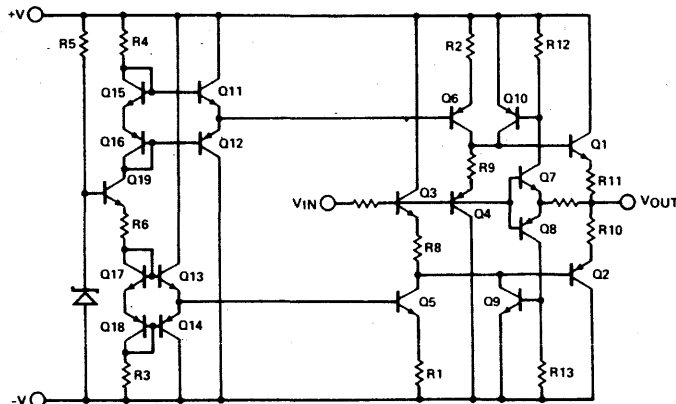
#### METAL CAN PACKAGE

#### TOP VIEWS



#### MINI - DIP

### SCHEMATIC



Harris Semiconductor LINEAR

Preliminary

### FEATURES

- LOW OFFSET VOLTAGE                    0.3mV
- HIGH SLEW RATE                        50V/ $\mu$ s
- ADJUSTABLE/GATEABLE GAIN
- LOW OFFSET CURRENT                  120nA
- ADJUSTABLE POWER CONSUMPTION

### APPLICATIONS

- USE AS BUILDING BLOCKS FOR:
- MULTIPLEXERS
- SAMPLE AND HOLD CIRCUITS
- COMPARATORS
- VOLTAGE FOLLOWERS

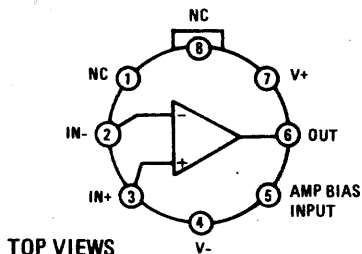
### DESCRIPTION

HA-23080 is an operational transconductance amplifier offering an unique blend of DC and AC electrical characteristics coupled with a bias input control for additional versatility. Fabricated using Harris Semiconductor's standard dielectric isolation process, these gateable-gain blocks provide input characteristics such as 0.3mV offset voltage, 120nA offset current and 150 $\mu$ V/V offset voltage sensitivity. Combined with these input characteristics are outstanding dynamic specifications including 50V/ $\mu$ s slew rate and 2MHz bandwidth.

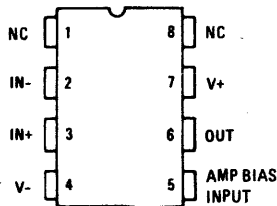
Using these features plus the flexibility afforded by the amplifier bias input, HA-23080 may be easily employed as the main component in multiplexer, comparator, multiplier, and sample/hold designs. This OTA is also useful as a fast voltage follower.

Using the popular operational amplifier pinout, HA-23080 comes available in 8 pin (TO-99) cans and 8 pin cerdips. Its specified temperature range is 0 $^{\circ}$ C to +75 $^{\circ}$ C for -5 and -55 $^{\circ}$ C to +125 $^{\circ}$ C for -2.

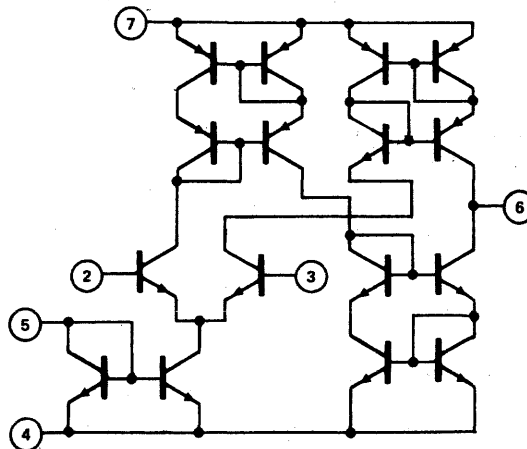
### PINOUT



TOP VIEWS



### SCHEMATIC





**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Voltage Between V+ and V- Terminals	36V	Power Dissipation	125mW
Differential Input Voltage	±5V	Operating Temperature Range	-55°C ≤ T <sub>A</sub> ≤ +125°C
Common Mode Input Voltage	V+ to V-	HA-23080-2	0°C ≤ T <sub>A</sub> ≤ +75°C
Input Signal Current	1mA	HA-23080-5	-65°C ≤ T <sub>A</sub> ≤ +150°C
Amplifier Bias Current	2mA	Storage Temperature Range	
Output Short Circuit Duration	Indefinite		

**ELECTRICAL CHARACTERISTICS** V+ = 15V, V- = -15V  
I<sub>ABC</sub> = 500μA (Unless Indicated Otherwise) (Note 2)

PARAMETER	TEMP.	HA-23080-2/-5			UNITS
		MINIMUM	TYPICAL	MAXIMUM	
<b>INPUT CHARACTERISTICS</b>					
Offset Voltage - I <sub>ABC</sub> = 5μA	+25°C		0.3	2	mV
	Full		0.4	5	mV
Offset Voltage Change I <sub>ABC</sub> = 5μA to I <sub>ABC</sub> = 500μA	+25°C		0.1	3	mV
Average Offset Voltage Drift - I <sub>ABC</sub> = 100μA	Full		3	18	μV/°C
Bias Current	+25°C		2	5	μA
	Full			8	μA
I <sub>ABC</sub> = 500nA	+25°C		2	5	nA
Offset Current	+25°C		0.12	0.6	nA
Common Mode Range	+25°C	±12	±13.6		V
Differential Input Current I <sub>ABC</sub> = 0, V <sub>DIFF</sub> = 4V	+25°C		0.008	5	nA
Leakage Current (See Test Circuit) I <sub>ABC</sub> = 0, V <sub>TP</sub> = 0V	+25°C		0.08	5	nA
	+25°C		0.3	5	nA
I <sub>ABC</sub> = 0, V <sub>TP</sub> = 36V					
Input Capacitance (Note 3)	+25°C		3.6		pF
Input Resistance	+25°C	10	26		kΩ
<b>TRANSFER CHARACTERISTICS</b>					
Forward Transconductance (Large Signal)	+25°C	7700	9900	15000	μmho
	Full	4000			μmho
Common Mode Rejection Ratio	+25°C	80	110		dB
Open Loop Bandwidth	+25°C		20		MHz
<b>OUTPUT CHARACTERISTICS</b>					
Output Voltage Swing (Note 4) - I <sub>ABC</sub> = 5μA	+25°C	±12	±13.8		V
	+25°C	±12	±13.5		V
Output Current (Note 5) - I <sub>ABC</sub> = 5μA	+25°C	3	5	7	μA
	+25°C	350	500	750	μA
	Full	300			μA
Output Capacitance (Note 3)	+25°C		5.6		pF
Output Resistance	+25°C		15		MΩ
Input-to-Output Capacitance (Note 3)	+25°C		0.024		pF
<b>TRANSIENT RESPONSE (See Test Circuit)</b>					
Slew Rate - Compensated	+25°C		50		V/μs
	+25°C		75		V/μs
<b>NOISE CHARACTERISTICS</b>					
Total Input Noise Voltage (Note 6)	+25°C			12	μVPEAK
<b>POWER SUPPLY CHARACTERISTICS</b>					
Supply Current	+25°C	0.8	1	1.5	mA
Input Offset Voltage Sensitivity	+25°C			150	μV/V
Amplifier Bias Voltage	+25°C		0.71		V

**NOTES:**

- Absolute maximum ratings are limiting values applied individually beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
- I<sub>ABC</sub> (Amplifier Bias Current) The current supplied to the amplifier bias terminal (Pin 5) to establish its operating point.
- f = 1MHz
- R<sub>L</sub> = ∞
- R<sub>L</sub> = 0
- V<sub>S</sub> = ±8.5V, I<sub>ABC</sub> = 500nA, Bandwidth = 1kHz, R<sub>SOURCE</sub> = 100kΩ, Test Time = 30 sec.

## SLIC-LC Subscriber Line Interface Circuit

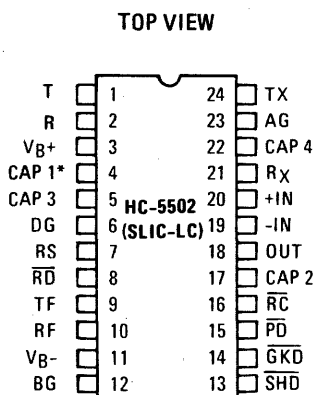
### FEATURES

- MONOLITHIC INTEGRATED DEVICE
- UNIQUE DI HIGH VOLTAGE PROCESS
- COMPATIBLE WITH WORLDWIDE PABX PERFORMANCE REQUIREMENTS
- CONTROLLED SUPPLY OF BATTERY FEED CURRENT FOR SHORT LOOPS
- INTERNAL RING RELAY DRIVER
- LOW POWER CONSUMPTION DURING STANDBY
- SWITCH HOOK, GROUND KEY AND RING TRIP DETECTION FUNCTIONS
- SELECTIVE DENIAL OF POWER TO SUBSCRIBER LOOPS

### APPLICATIONS

- SOLID STATE LINE INTERFACE CIRCUIT FOR ANALOG AND DIGITAL PBX SYSTEMS

### PINOUT



\*Optional

### DESCRIPTION

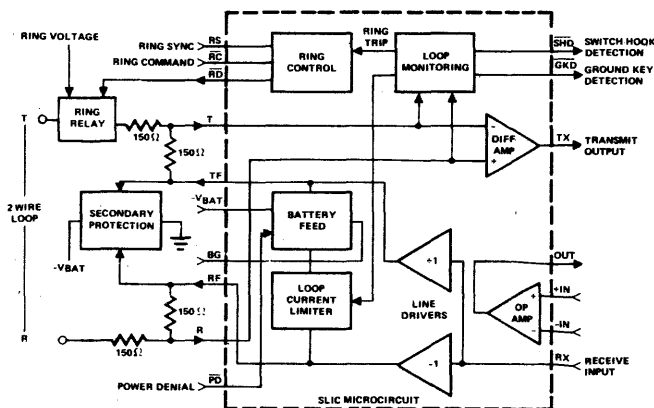
The HARRIS SLIC-LC incorporates many of the BORSHT functions on a single IC chip. This includes DC battery feed, a ring relay driver, supervisory and hybrid functions. Using the unique HARRIS dielectric isolation process, the SLIC-LC can operate directly with a wide range of station battery voltages.

The SLIC-LC also provides selective denial of power. If the PABX system becomes overloaded during an emergency, the SLIC-LC will provide system protection by denying power to selected subscriber loops.

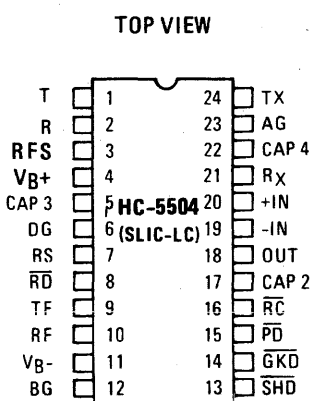
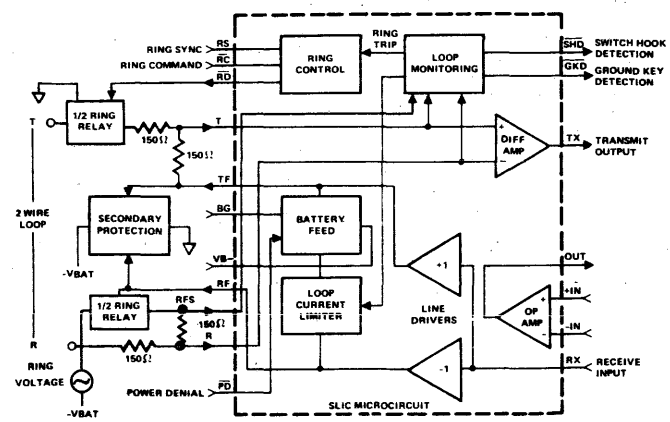
The HARRIS SLIC-LC is ideally suited in the design of new digital PABX systems, by eliminating bulky, expensive hybrid transformers.

SLIC-LC is available in either a 24 pin dual-in-line plastic or ceramic package.

### FUNCTIONAL DIAGRAM



## SLIC-LC Subscriber Line Interface Circuit

FEATURES	DESCRIPTION																																																
<ul style="list-style-type: none"> <li>● MONOLITHIC INTEGRATED DEVICE</li> <li>● UNIQUE DI HIGH VOLTAGE PROCESS</li> <li>● COMPATIBLE WITH WORLDWIDE PABX PERFORMANCE REQUIREMENTS</li> <li>● CONTROLLED SUPPLY OF BATTERY FEED CURRENT FOR SHORT LOOPS</li> <li>● INTERNAL RING RELAY DRIVER</li> <li>● ALLOWS INTERFACING WITH NEGATIVE SUPERIMPOSED RINGING SYSTEMS</li> <li>● LOW POWER CONSUMPTION DURING STANDBY</li> <li>● SWITCH HOOK, GROUND KEY AND RING TRIP DETECTION FUNCTIONS</li> <li>● SELECTIVE DENIAL OF POWER TO SUBSCRIBER LOOPS</li> </ul>	<p>The HARRIS SLIC-LC incorporates many of the BORSHT functions on a single IC chip. This includes DC battery feed, a ring relay driver, supervisory and hybrid functions. Using the unique HARRIS dielectric isolation process, the SLIC-LC can operate directly with a wide range of station battery voltages.</p> <p>The SLIC-LC also provides selective denial of power. If the PABX system becomes overloaded during an emergency, the SLIC-LC will provide system protection by denying power to selected subscriber loops.</p> <p>The HARRIS SLIC-LC is ideally suited for the design of new PABX systems, by eliminating bulky, expensive hybrid transformers.</p> <p>SLIC-LC is available in either a 24 pin dual-in-line plastic or ceramic package. The SLIC-LC is also available in die form.</p>																																																
<h3>APPLICATIONS</h3>																																																	
<ul style="list-style-type: none"> <li>● SOLID STATE LINE INTERFACE CIRCUIT FOR ANALOG AND DIGITAL PBX SYSTEMS</li> </ul>																																																	
PINOUT	FUNCTIONAL DIAGRAM																																																
<p style="text-align: center;">TOP VIEW</p>  <table border="1" style="width: 100%; border-collapse: collapse; text-align: center;"> <tr> <td>T</td><td>1</td><td>24</td><td>TX</td> </tr> <tr> <td>R</td><td>2</td><td>23</td><td>AG</td> </tr> <tr> <td>RFS</td><td>3</td><td>22</td><td>CAP 4</td> </tr> <tr> <td>VB+</td><td>4</td><td>21</td><td>RX</td> </tr> <tr> <td>CAP 3</td><td>5</td><td>20</td><td>+IN</td> </tr> <tr> <td>DG</td><td>6 (SLIC-LC)</td><td>19</td><td>-IN</td> </tr> <tr> <td>RS</td><td>7</td><td>18</td><td>OUT</td> </tr> <tr> <td>RD</td><td>8</td><td>17</td><td>CAP 2</td> </tr> <tr> <td>TF</td><td>9</td><td>16</td><td>RC</td> </tr> <tr> <td>RF</td><td>10</td><td>15</td><td>PD</td> </tr> <tr> <td>VB-</td><td>11</td><td>14</td><td>GKD</td> </tr> <tr> <td>BG</td><td>12</td><td>13</td><td>SHD</td> </tr> </table>	T	1	24	TX	R	2	23	AG	RFS	3	22	CAP 4	VB+	4	21	RX	CAP 3	5	20	+IN	DG	6 (SLIC-LC)	19	-IN	RS	7	18	OUT	RD	8	17	CAP 2	TF	9	16	RC	RF	10	15	PD	VB-	11	14	GKD	BG	12	13	SHD	 <p>The functional diagram illustrates the internal architecture of the SLIC-LC microcircuit. It shows a 2-wire loop connected to a 1/2 ring relay. The circuit includes a secondary protection stage, a battery feed section with a loop current limiter, and line drivers. Key control and monitoring blocks include Ring Control, Loop Monitoring, and a differential amplifier (DIFF AMP). The diagram also shows connections for Ring Sync (RS), Ring Command (RC), Ring Trip, Loop Trip, Switch Hook Detection (SHD), Ground Key Detection (GKD), Transmit Output (TX), Receive Input (RX), and various power and control signals like VBAT, R, RFS, RF, TF, BG, and PD.</p>
T	1	24	TX																																														
R	2	23	AG																																														
RFS	3	22	CAP 4																																														
VB+	4	21	RX																																														
CAP 3	5	20	+IN																																														
DG	6 (SLIC-LC)	19	-IN																																														
RS	7	18	OUT																																														
RD	8	17	CAP 2																																														
TF	9	16	RC																																														
RF	10	15	PD																																														
VB-	11	14	GKD																																														
BG	12	13	SHD																																														



**HARRIS**

# HC-5512/5512A

## PCM Monolithic Filter

### FEATURES

- EXCEEDS ALL D3/D4 AND CCITT SPECIFICATIONS
- +5V, -5V POWER SUPPLIES
- LOW POWER CONSUMPTION:
  - 45mW (600Ω 0dBm LOAD)
  - 30mW (POWER AMPS DISABLED)
- POWER DOWN MODE: 0.5mW
- 20dB GAIN ADJUST RANGE
- NO EXTERNAL ANTI-ALIASING COMPONENTS
- SIN x/x CORRECTION IN RECEIVE FILTER
- 50/60Hz REJECTION IN TRANSMIT FILTER
- TTL AND CMOS COMPATIBLE LOGIC
- ALL INPUT PROTECTED AGAINST STATIC DISCHARGE DUE TO HANDLING

### DESCRIPTION

The HC-5512/HC-5512A filter is a monolithic circuit containing both transmit and receive filters specifically designed for PCM CODEC filtering applications in 8kHz sampled systems.

The filter is manufactured using double-poly silicon gate CMOS technology. Switched capacitor integrators are used to simulate classical LC ladder filters which exhibit low component sensitivity.

#### TRANSMIT FILTER STAGE

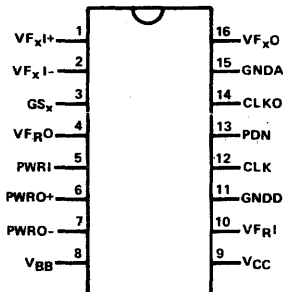
The transmit filter is a fifth order elliptic low pass filter in series with a fourth order Chebyshev high pass filter. It provides a flat response in the passband and rejection of signals below 200Hz and above 3.4kHz.

#### RECEIVE FILTER STAGE

The receive filter is a fifth order elliptic low pass filter designed to reconstruct the voice signal from the decoded/demultiplexed signal which, as a result of the sampling process, is a stair-step signal having the inherent sin x/x frequency response. The receive filter approximates the function required to compensate for the degraded frequency response and restore the flat passband response.

### PINOUT

DUAL-IN-LINE PACKAGE  
TOP VIEW



### FUNCTIONAL DIAGRAM

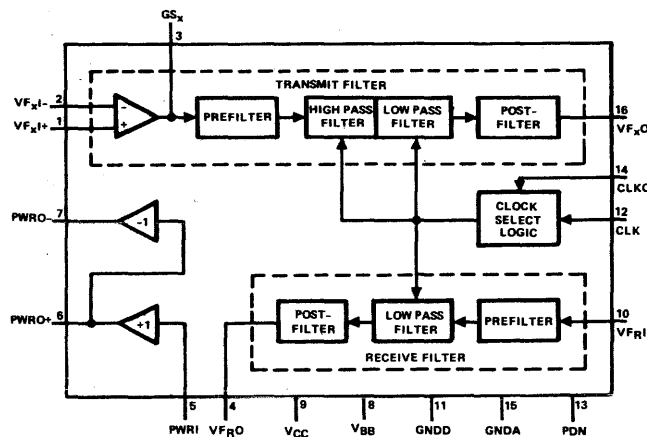


FIGURE 1

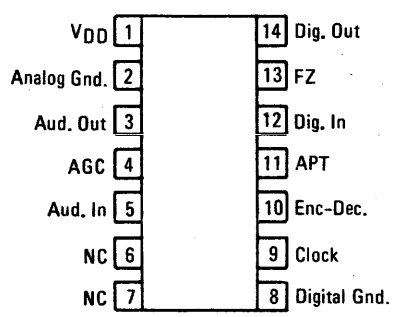
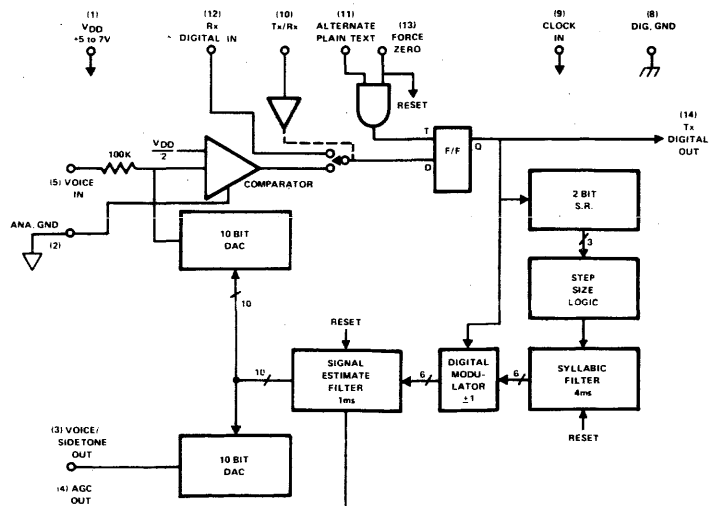
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## All-Digital Continuously Variable Slope Delta Modulator (CVSD)

*Preliminary*

FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>REQUIRES FEW EXTERNAL PARTS</li> <li>LOW POWER DRAIN: 6mW FROM SINGLE 4.5V-7V SUPPLY</li> <li>TIME CONSTANTS DETERMINED BY CLOCK FREQUENCY; NO CALIBRATION OR DRIFT PROBLEMS; AUTOMATIC OFFSET ADJUSTMENT</li> <li>HALF DUPLEX OPERATION UNDER DIGITAL CONTROL</li> <li>FILTER RESET UNDER DIGITAL CONTROL</li> <li>AUTOMATIC OVERLOAD RECOVERY</li> <li>AUTOMATIC "QUIET" PATTERN GENERATION</li> <li>AGC CONTROL SIGNAL AVAILABLE</li> </ul>	<p>The HC-55564 is a half duplex modulator/demodulator CMOS integrated circuit used to convert voice signals into serial NRZ digital data, and to reconvert that data into voice. The conversion is by delta modulation, using the continuously variable slope (CVSD) method of companding.</p> <p>While the signals are compatible with other CVSD circuits, the internal design is unique. The analog loop filters have been replaced by digital filters, using very low power, and requiring no external timing components. This approach allows inclusion of many desirable features which would be difficult to implement using other approaches.</p> <p>The HC-55564 is usable from 9K bits/sec to above 64K bits/sec. The unit is available in a 14 pin ceramic or plastic DIP package in commercial and military temperature ranges including MIL-STD 883-B processing.</p>
APPLICATIONS	
<ul style="list-style-type: none"> <li>VOICE TRANSMISSION OVER DATA CHANNELS</li> <li>VOICE ENCRYPTION/SCRAMBLING</li> <li>VOICE I/O FOR DIGITAL SYSTEMS AND SPEECH SYNTHESIS</li> <li>AUDIO MANIPULATIONS: DELAY LINES, TIME COMPRESSION, ECHO GENERATION/SUPPRESSION, SPECIAL EFFECTS, ETC.</li> </ul>	

PINOUT	FUNCTIONAL DIAGRAM
<p><b>Top View</b></p> 	

## All-Digital Continuously Variable Slope Delta Demodulator (CVSD)

### FEATURES

- REQUIRES FEWER EXTERNAL PARTS
- LOW POWER DRAIN: 6mW FROM SINGLE 5V-7V SUPPLY
- TIME CONSTANTS DETERMINED BY CLOCK FREQUENCY; NO CALIBRATION OR DRIFT PROBLEMS; AUTOMATIC OFFSET ADJUSTMENT
- FILTER RESET BY DIGITAL CONTROL
- AUTOMATIC OVERLOAD RECOVERY
- AUTOMATIC "QUIET" PATTERN GENERATION

### APPLICATIONS

- SPEECH SYNTHESIS
- AUDIO MANIPULATIONS; DELAY LINES, ECHO GENERATION/SUPPRESSION, SPECIAL EFFECTS, ETC.

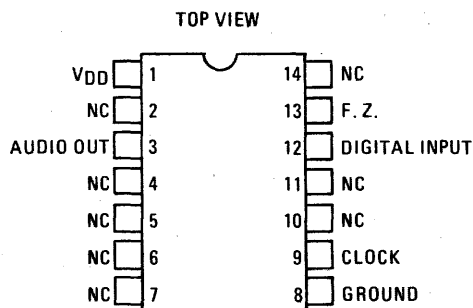
### DESCRIPTION

The HC-55536 is a CMOS integrated circuit used to convert serial NRZ digital data to an analog (voice) signal. Conversion is by delta demodulation, using the continuously variable slope (CVSD) method.

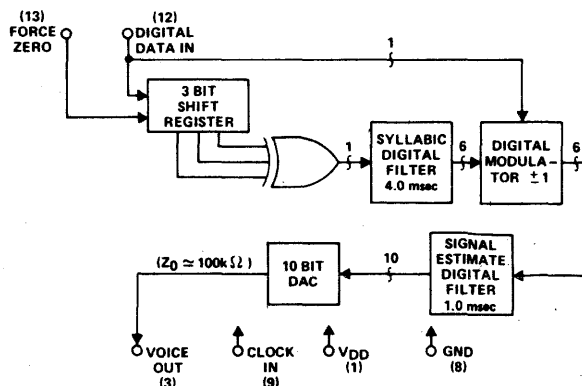
While signals are compatible with other CVSD circuits, the internal design is unique. The analog loop filters have been replaced by digital filters which use very low power and require no external timing components. This digital approach allows inclusion of many desirable features, which otherwise would be difficult to implement. Internal time constants are optimized for a 16K bit/sec data rate. However, the device is usable from 9K bits/sec to above 64K bits/sec.

The package is a 14 pin DIP, available in plastic (0°C to +70°C) or ceramic (-40°C to +85°C). Chips are available, probe tested at +25°C.

### PINOUT AND PIN DESCRIPTION

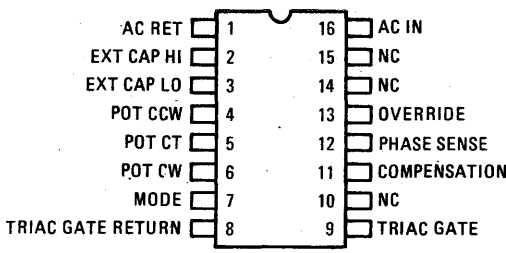


### FUNCTIONAL DIAGRAM

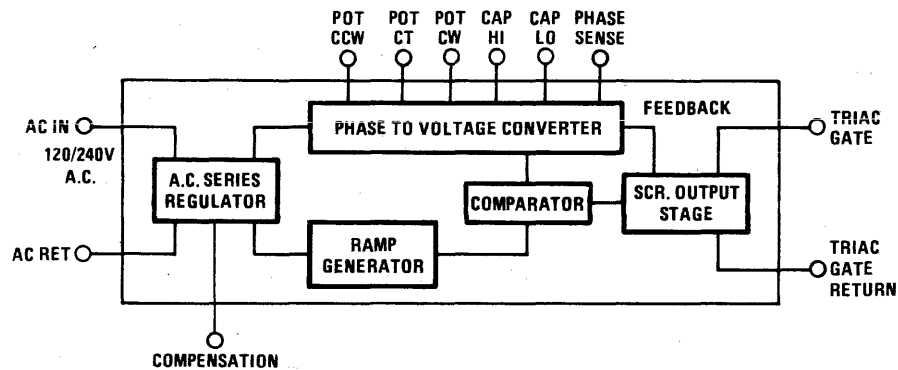


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FEATURES	DESCRIPTION
<ul style="list-style-type: none"> <li>● OPERATES DIRECTLY OFF 120V/240V AC LINE – NO POWER SUPPLY REQUIRED</li> <li>● 50Hz OR 60Hz OPERATION</li> <li>● PRODUCES POWER SAVINGS FROM 10% TO 50% FOR MOTORS WITH LIGHT OR VARIABLE LOADS</li> <li>● SCR OUTPUT STAGE TRIGGERS TRIAC DIRECTLY</li> <li>● LOAD ANTICIPATOR SENSES SHOCK LOADS AND RESPONDS INSTANTLY WITH FULL POWER</li> <li>● WITHSTANDS LINE SURGES TO 3500V</li> <li>● CAUSES MOTOR TO RUN QUIETER, COOLER</li> <li>● CAN BE MOUNTED INSIDE MOTOR</li> <li>● NEEDS ONLY 3 RESISTORS, 3 CAPACITORS AND A TRIAC TO ASSEMBLE COMPLETE CONTROLLER</li> </ul>	<p>The HV-1000 and HV-1000A are energy saving induction motor control circuits designed to reduce the power consumed by single phase induction motors. HV-1000 is for 120V AC use and HV-1000A for 240V AC use.</p> <p>The controller circuit senses the load on the motor and then controls a TRIAC to apply reduced voltage to lightly loaded motors, full voltage to heavily loaded motors.</p> <p>The HV-1000/1000A are available in a 16 lead plastic DIP. Ideal for mounting inside induction motors, they can also be mounted in a heat sunk circuit box for external, after market application.</p>
APPLICATIONS	PINOUT
<ul style="list-style-type: none"> <li>● MACHINE TOOLS</li> <li>● INDUSTRIAL SEWING MACHINES</li> <li>● HEAT PUMPS</li> <li>● PRESSES</li> <li>● CONVEYORS</li> <li>● DISC PACK DRIVES</li> </ul> <p>▶ ANY APPLICATION WHERE FOR SOME OF THE TIME THE MOTOR IS DRIVING LESS THAN ITS RATED LOAD</p>	<p style="text-align: center;">TOP VIEW</p> 

### FUNCTIONAL DIAGRAM



**Operational Amplifiers — General Purpose**

Type	Description	V <sub>OS</sub> (mV)	I <sub>b</sub> (nA)	A <sub>VOL</sub> (V/V)	GBW (typ) (MHz)	I <sub>SUPP</sub> (mA)	T <sub>A</sub> (°C)	Packages*	Remarks
108	Low Level, Uncompensated	2.0	2.0	50,000	1.0	0.6	-55, +125	J, F, T	
108LN	Guaranteed Noise 108	2.0	2.0	50,000	1.0	0.6	-55, +125	T	70nV/√Hz @10Hz
308	Low Level, Uncompensated	7.5	7.0	25,000	1.0	0.8	0, +70	F, J, P, T	
308LN	Guaranteed Noise 308	7.5	7.0	25,000	1.0	0.8	0, +70	T	70nV/√Hz @10Hz
777	General Purpose Comparator	0.7	25	150,000	0.8	2.5	-55, +125	P, T	
777C	General Purpose Comparator	0.7	25	150,000	0.8	2.5	0, +70	P, T	
8008M	Low Bias Current, Compensated	5.0	10	20,000	1.0	2.8	-55, +125	J, T	
8008C	Low Bias Current, Compensated	6.0	25	20,000	1.0	2.8	0, +70	J, P, T	
LH2108	Dual Super Beta	2.0	3.0	25,000	1.0	0.4	-55 to +125	D	} Build to Order
LH2108A	Dual Super Beta	0.5	3.0	40,000	1.0	0.4	-55 to +125	D	
LH2308	Dual Super Beta	7.5	10	15,000	1.0	0.4	0 to +70	D	
LH2308A	Dual Super Beta	0.5	10	60,000	1.0	0.4	0 to +70	D	

**Operational Amplifiers — Low Power Programmable**

Type	Description	V <sub>OS</sub> (mV)	I <sub>b</sub> (nA)	A <sub>VOL</sub> (V/V)	GBW (MHz)	I <sub>SUPP</sub> (μA)	@ I <sub>sat</sub> (μA)	@ V <sub>s</sub> (V)	T <sub>A</sub> (°C)	Packages*
8021M	Programmable, Compensated	3.0	20	50,000	0.27	40	30	+6.0	-55 to +125	J, T
8021C	Programmable, Compensated	6.0	30	50,000	0.27	50	30	+6.0	0, +70	T
8022M	Dual 8021M	3.0	20	50,000	0.27	40	30	+6.0	-55 to +125	J, F
8022C	Dual 8021C	6.0	30	50,000	0.27	50	30	+6.0	0, +70	J, P
8023M	Triple 8021M	3.0	20	50,000	0.27	40	30	+6.0	-55 to +125	J
8023C	Triple 8021C	6.0	30	50,000	0.27	50	30	+6.0	0 to +70	J, P
7611	CMOS	2.0	0.001	100,000	1.4	20	-	-	C, I, M	T, P
7612	CMOS, Extended CMVR	2.0	0.001	100,000	1.4	20	-	-	C, I, M	T, P
7613	CMOS, Input Protected to ±200V	2.0	0.001	100,000	1.4	20	-	-	C, I, M	T, P
7631	CMOS, Triple	5.0	0.001	100,000	1.4	60	-	-	C, I, M	D, P
7632	CMOS, Triple, Uncompensated	5.0	0.001	100,000	1.4	60	-	-	C, I, M	D, P

**Operational Amplifiers — CMOS**

Type	Description	Compensation	Offset Null	V <sub>OS</sub> Selection	I <sub>OS</sub>	I <sub>b</sub>	Output Swing	Input CMR	Packages*
7611	Single, Selectable I <sub>O</sub>	Internal	Yes	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, T
7612	Single, Selectable I <sub>O</sub> Extended CMVR	Internal	Yes	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> +300mV	V <sub>SUPP</sub> -100mV	P, T
7613	Single, Selectable I <sub>O</sub> Input Protected to ±200V	Internal	Yes	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, T
7614	Single, Fixed I <sub>O</sub>	External	Yes	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, T
7615	Single, Fixed I <sub>O</sub> Input Protected	External	Yes	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, T
7621	Dual, Fixed I <sub>O</sub>	Internal	No	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, T
7622	Dual, Fixed I <sub>O</sub>	Internal	Yes	2, 5, 15mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, J
7631	Triple, Selectable I <sub>O</sub>	Internal	No	5, 10, 20mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, J
7632	Triple, Selectable I <sub>O</sub>	None	No	5, 10, 20mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, J
7641	Quad, Fixed I <sub>O</sub>	Internal	No	5, 10, 20mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, J
7642	Quad, Fixed I <sub>O</sub>	Internal	No	5, 10, 20mV	0.5pA	1pA	V <sub>SUPP</sub> -100mV	V <sub>SUPP</sub> -100mV	P, J
7650	Chopper Stabilized	Internal	-	0.01mV	-	10pA	V <sub>SUPP</sub> -100mV	-	P, T

\*Package Key: D—Solder lid side brazed ceramic dual-in-line. F—Ceramic flat package. J—Glass frit seal ceramic dual-in-line. P—Plastic dual-in-line. T—Metal can.

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### Operational Amplifiers — FET Input (also see Operational Amplifier, CMOS)

Type	Description	V <sub>OS</sub> (mV)	I <sub>b</sub> ( $\mu$ A)	A <sub>VOL</sub> (V/V)	GBW (typ) (MHz)	Slew		T <sub>A</sub> (°C)	Packages*	Remarks
						Rate (V/ $\mu$ s)	I <sub>SUPP</sub> (mA)			
8007M	General Purpose, Compensated	20	20	50,000	1.0	6	5.2	-55, +125	T	
8007AM	8007M, Low I <sub>b</sub>	30	1.0	20,000	1.0	2.5	6	-55, +125	T	
8007C	General Purpose, Compensated	50	50	20,000	1.0	6	6	0, +70	T	All BIFET amplifiers offer low noise — see data sheets
8007AC	8007C, Low I <sub>b</sub>	30	1.0	20,000	1.0	2.5	6	0, +70	T	
8043M	Dual 8007M	20	20	50,000	1.0	6.0	6	-55, +125	J	
8043C	Dual 8007C	50	50	20,000	1.0	6.0	6.8	-55, +125	J, P	
8500	MOSFET Input, Compensated	50	0.1	20,000	0.7	0.5	2.7	-25, +85	T	
8500A	MOSFET Input, Super Low I <sub>b</sub>	50	0.01	20,000	0.7	0.5	2.7	-25, +85	T	

### Operational Amplifiers — High Performance

Type	Description	V <sub>OS</sub> (mV)	I <sub>b</sub> (pA)	A <sub>VOL</sub> (V/V)	GBW (MHz)	Slew		T <sub>A</sub> (°C)	Packages*
						Rate (V/ $\mu$ s)	I <sub>SUPP</sub> (mA)		
8017M	High Speed, Inverting	5.0	200	25,000	10	130	7.0	-55, +125	T, F
8017C	High Speed, Inverting	7.0	200	25,000	10	130	8.0	0, +70	T, F
OP-05	Low Bias, Low Drift	0.07	700	500,000	0.6	0.2	4.0	-55, +125	T, J
OP-07	Ultra Stable	0.025	300	500,000	0.6	0.17	4.0	-55, +125	T, J

### Operational Amplifiers—High Slew Rate

Type	Description	V <sub>OS</sub> (mV)	I <sub>b</sub> (pA)	A <sub>VOL</sub> (V/V)	GBW (MHz)	Slew		T <sub>A</sub> (°C)	Packages*
						Rate (V/ $\mu$ s)	I <sub>SUPP</sub> (mA)		
8017M	High speed, inverting	5.0	200	25,000	10	130	7.0	-55 - 125	T, F
8017C	High speed, inverting	7.0	200	25,000	10	130	8.0	0 - 70	T, F

\*Package Key: D—Solder lid side brazed ceramic dual-in-line. F—Ceramic flat package. J—Glass frit seal ceramic dual-in-line. P—Plastic dual-in-line. T—Metal can.

### Precision Operational Amplifiers, $V_{SUPP} = \pm 2V$ to $\pm 8V$

Type	Description	$V_{OS}$ ( $\mu V$ )	$\Delta V_{OS}$ ( $\mu V/^{\circ}C$ )	$\Delta V_{OS}$ ( $\mu V/year$ )	$A_V$ (dB min)	Slew Rate Rate (V/ $\mu s$ )	$I_{BIAS}$ ( $\mu A$ )	Packages*	$T_A$ ( $^{\circ}C$ )
ICL7650C	Chopper Stabilized	$\pm 1$	$\pm 0.01$	$100nV/\sqrt{month}$	126	2.5	+1.5	J, P, T	0 to +70
ICL7650I	Chopper Stabilized	$\pm 1$	$\pm 0.01$	$100nV/\sqrt{month}$	126	2.5	+1.5	J, P, T	-25 to +85
ICL7652	Chopper Stabilized	$\pm 1$	$\pm 0.7$	$100nV/\sqrt{month}$	120	0.5	+1.5	J, P, T	-20 to +85

### Precision Instrumentation Amplifiers, $V_{SUPP} = \pm 2V$ to $\pm 5V$ , $I_{SUPP} = 1.7mA$

Type	Description	$V_{OS}$ ( $\mu V$ )	$\Delta V_{OS}$ ( $\mu V/^{\circ}C$ )	$\Delta V_{OS}$ ( $\mu V/year$ )	$A_V$ (dB min)	Packages*	$T_A$ ( $^{\circ}C$ )
ICL7605C	Compensated	$\pm 2$	$\pm 0.01$	0.5	90	J, P	0 to +70
ICL7605I	Compensated	$\pm 2$	$\pm 0.01$	0.5	90	J, P	-25 to +85
ICL7605M	Compensated	$\pm 2$	$\pm 0.05$	0.5	90	J, P	-55 to +125
ICL7606C	Uncompensated	$\pm 2$	$\pm 0.01$	0.5	90	J, P	0 to +70
ICL7606I	Uncompensated	$\pm 2$	$\pm 0.01$	0.5	90	J, P	-25 to +85
ICL7606M	Uncompensated	$\pm 2$	$\pm 0.05$	0.5	90	J, P	-55 to +125

### Precision Voltage References

Type	Description	
ICL8069	Low Voltage Reference	The ICL8069 is a 1.2V temperature compensated voltage reference. It uses the band-gap principle to achieve excellent stability and low noise at reverse currents down to $50\mu A$ .
ICL8075-9	Ultra Precision Temperature Stabilized Voltage References	The ICL8075-9 is a family of precision laser-trimmed voltage references that incorporate a substrate heater to produce extremely low overall voltage temperature coefficients. The series of devices is produced so that exact voltages are available for the most popular A/D and D/A converters. This avoids the necessity to perform adjustments in most cases, and reduces the problems with trim range and temperature coefficient loss in all others.

### Video Amplifiers

Type	Description	Gains (typ) (V/V)	Bandwidths (typ) (MHz)	$e_n$ $\mu V(rms)$	Output Offset (V)	$I_{SUPP}$ (mA)	$T_A$ ( $^{\circ}C$ )	Packages*
NE/SE592	Gain Selectable Video Amp	400, 100, 10	40, 90	12	0.75	10	0, +70/-55, +125	J, T
NE592-8	Gain Selectable Video Amp	400, 100, 10	40	12	0.75	10	0, +70	P

\*Package Key: D—Solder lid side brazed ceramic dual-in-line. F—Ceramic flat package. J—Glass frit seal ceramic dual-in-line. P—Plastic dual-in-line. T—Metal can.



## Comparators

Type	Description	V <sub>OS</sub> (mV)	I <sub>b</sub> (nA)	A <sub>v</sub> (V/mV)	t <sub>pd</sub> (ns)(typ)	I <sub>SUPP</sub> (mA)	V <sub>OL</sub> (V)	I <sub>OL</sub> (mA)	T <sub>A</sub> (°C)	Packages*
8001M	Low Power Comparator	3	100	15	250	2	0.5	2	-55, +125	T
8001C	Low Power Comparator	5	250	15	250	2	0.4	2	0, +70	T

Notes: t<sub>pd</sub> measured for 100mV step with 5mV overdrive.

I<sub>SUPP</sub> measured for V<sub>SUPP</sub> + ± 15V.

## Power Amplifiers

Type	Description	Use	Output Current (A)	Output Swing (V)	V <sub>OS</sub> (mV)	I <sub>b</sub> (nA)	A <sub>VOL</sub> (V/V)	Slow Rate (V/μs)	Quiescent I <sub>SUPP</sub> (mA)	T <sub>A</sub> (°C)
ICH8510M	Hybrid Power Amplifier		1.0	± 26	3.0	250	100,000	0.5	40	-55, +125
ICH8510I	Hybrid Power Amplifier		1.0	± 26	6.0	500	100,000	0.5	50	-25, +85
ICH8515I	Hybrid Power Amplifier		1.25	± 12	6.0	500	100,000	0.5	80	-20°C - +85°C
ICH8515M	Hybrid Power Amplifier	Servo	1.5	± 12	3.0	250	100,000	0.5	70	-55°C - +125°C
ICH8520M	Hybrid Power Amplifier	and	2.0	± 26	3.0	250	100,000	0.5	40	-55, +125
ICH8520I	Hybrid Power Amplifier	Actuator	2.0	± 26	6.0	500	100,000	0.5	50	-25, +85
ICH8530M	Hybrid Power Amplifier		2.7	± 25	3.0	250	100,000	0.5	40	-55, +125
ICH8530I	Hybrid Power Amplifier	Power	2.7	± 25	6.0	500	100,000	0.5	50	-25, +85
ICL8063C	Monolithic Power Amplifier	Transistors	2.0	± 27	50		6		250	0, +70
ICL8063M	Monolithic Power Amplifier		2.0	± 27	75		6		300	-55, +125

Note 1: Specifications apply at ± 30V supplies.

Note 3: Fully protected against inductive current flow.

Note 2: All units packaged in 8 lead TO-3 can.

Note 4: Externally settable output current limiting.

\*Package Key: D—Solder lid side brazed ceramic dual-in-line. F—Ceramic flat package. J—Glass frit seal ceramic dual-in-line. P—Plastic dual-in-line.  
T—Metal can.

## Special Function Circuits

Type	Description	Accuracy	V <sub>supp</sub> (V)	T <sub>A</sub> (°C)	Packages*
AD590	Temperature transducer—output linear at 1 $\mu$ A/°K	$\pm 1^\circ\text{C}$	40 to 30	-55 to +150	F, H
ICL7667C	Dual Power MOS Driver	—	22	0, +70	J, P, T
ICL7667M	Dual Power MOS Driver	—	22	-55, +125	J, T
8013AM	Four quadrant multiplier. Output proportional to algebraic products	$\pm 0.5\%$	$\pm 15$	-55, +125	T
8013BM	of two input signals. Features $\pm 0.5\%$ accuracy; internal op amp	$\pm 1.0\%$	$\pm 15$	-55, +125	T
8013CM	for level shift, division and square root functions; full $\pm 10\text{V}$	$\pm 2.0\%$	$\pm 15$	-55, +125	T
8013AC	input/output range; 1MHz bandwidth.	$\pm 0.5\%$	$\pm 15$	0, +70	T
8013BC		$\pm 1.0\%$	$\pm 15$	0, +70	T
8013CC		$\pm 2.0\%$	$\pm 15$	0, +70	T
8038AM	Simultaneous Sine, Square, and Triangle wave outputs T <sup>2</sup> L com-	1.5%	$\pm 5$ to $\pm 15$	-55, +125	J
8038AC	patible to 28V over frequency range from 0.01Hz to 1.0MHz.	1.5%	$\pm 5$ to $\pm 15$	0, +70	J
8038BM	Low distortion (< 1%); high linearity (0.1%); low frequency drift	3.0%	$\pm 5$ to $\pm 15$	-55, +125	J
	with temperature (50ppm/°C max); variable duty cycle (2%-98%).				
8038BC		3.0%	$\pm 5$ to $\pm 15$	0, +70	P
8038CC	External frequency modulation.	5.0%	$\pm 5$ to $\pm 15$	0, +70	P
8048BC	Log amp 1V/decade (Adjustable). 120dB range.	$\pm 30\text{mV}$	$\pm 15$	0, +70	J, P
8048CC	with current input. Error referred to output.	$\pm 60\text{mV}$	$\pm 15$	0, +70	J, P
8049BC	Antilog amplifier adjustable scale factor.	$\pm 10\text{mV}$	$\pm 15$	0, +70	J, P
8049CC	Error referred to input.	$\pm 30\text{mV}$	$\pm 15$	0, +70	J, P
8211M	Micropower voltage detector/indicator/voltage regulator/		2 to 30	-55, +125	T
8211C	programmable zener. Contains 1.15V micropower reference		2 to 30	0, +70	P, T
8212M	plus comparator and hysteresis output. Main output		2 to 30	-55, +125	T
8212C	inverting (8212) or non-inverting (8211).		2 to 30	0, +70	P, T

Note: All parameters are specified at V<sub>supp</sub> =  $\pm 15\text{V}$  and T<sub>A</sub> = +25°C unless otherwise noted.

\*Package Key: D—Solder lid side brazed ceramic dual-in-line. F—Ceramic flat package. J—Glass frit seal ceramic dual-in-line. P—Plastic dual-in-line. T—Metal can.

## CMOS Power Supply Circuits

Type	Description	
ICL7660	Voltage Converter	The ICL7660 performs the complete supply voltage conversion from positive to negative for an input range of +1.5V to +10.0V, resulting in complementary output voltages of -1.5V to -10.0V.
ICL7663	Positive Regulator	The ICL7663 (positive) and ICL7664 (negative) series regulators are low-power, high-efficiency devices which accept inputs from 1.6V to 16V and provide adjustable outputs over the same range at currents up to 40mA. Operating current is typically less than 40 $\mu$ A, regardless of load.
ICL7664	Negative Regulator	
ICL7665	Programmable Micropower Voltage Detector	The ICL7665 contains two individually programmable voltage detectors on a single chip. Requiring only $\sim 3\mu\text{A}$ for operation, the device is intended for battery-operated systems and instruments which require high or low voltage warnings, settable trip points, or fault monitoring and correction.



# ICH8500 / A

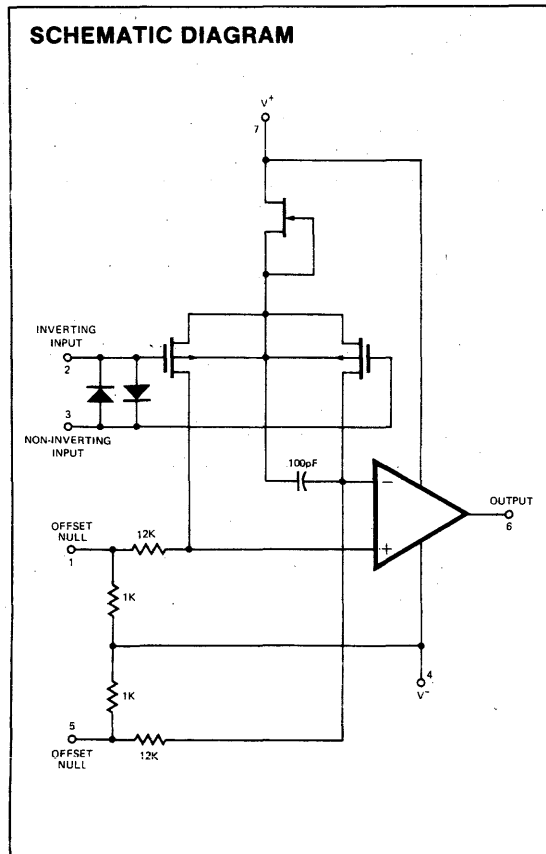
## Ultra Low Bias Current Operational Amplifier

### FEATURES

- Input diode protection
- Input bias current less than 0.01 pA at all operating temperatures
- No frequency compensation required
- Offset voltage null capability
- Short circuit protection
- Low power consumption

### APPLICATIONS

- Femto Ammeter
- Electrometers
- Long time integrators
- Flame detectors
- pH meter
- Proximity detector
- Sample and Hold Circuits



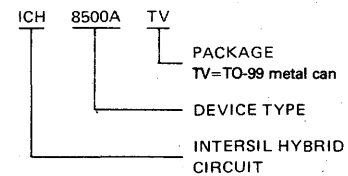
### GENERAL DESCRIPTION

The ICM8500 and ICH8500A are hybrid circuits designed for ultra low input bias current operational amplifier applications. They are ideally suited for analog and electrometer applications where high input resistance and low input current are of prime importance.

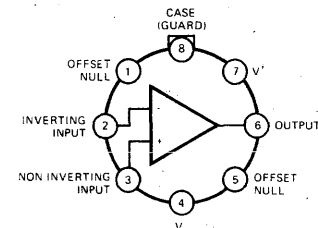
Functionally, they are pin for pin identical to the popular 741 monolithic amplifier. These amplifiers are unconditionally stable and the input offset voltage can be adjusted to zero with an external 20k potentiometer. The input bias current for the inverting and non-inverting inputs is 0.1 pA maximum for the ICH8500, and 0.01 pA maximum for the ICH8500A and are constant over the operating temperature range of -25°C to +85°C.

Pin 8 is connected to the case. This permits the designer to operate the case at any desired potential, the key to achieving the ultra low input currents associated with these two amplifiers. Forcing the case to the same potential as the inputs eliminates current flow between the case and the input pins, and leakage currents that may have otherwise existed between any of the other pins and the inputs are intercepted by the case.

### ORDERING INFORMATION



### PIN CONFIGURATION (outline dwg TV)



# ICH8500/A



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage .....  $\pm 18V$   
 Internal Power Dissipation<sup>(1)</sup> ..... 500 mW  
 Differential Voltage .....  $\pm 0.5V$   
 Storage Temperature .....  $-65^{\circ}C$  to  $+150^{\circ}C$   
 Operating Temperature .....  $-25^{\circ}C$  to  $+85^{\circ}C$   
 Lead Temperature (Soldering 10 sec) .....  $300^{\circ}C$   
 Output Short Circuit Duration ..... Indefinite

NOTE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent device failure. These are stress ratings only and functional operation of the devices at these or any other conditions above those indicated in the operation sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may cause device failures.

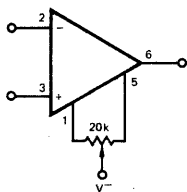
Note: 1. Rating applies for ambient temperature to  $+70^{\circ}C$ .

## ELECTRICAL CHARACTERISTICS ( $T_A = 25^{\circ}C$ unless otherwise specified, $V_{SUPP} = \pm 15V$ )

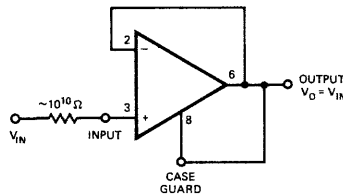
CHARACTERISTICS	SYMBOL	ICH8500			ICH8500A			UNITS	TEST CONDITIONS
		MIN	TYP.	MAX	MIN	TYP.	MAX		
Input Leakage Current (Inverting and Non-Inverting)	$I_{ILK}$			0.1			0.01	pA	Case at same potential as inputs
Input Offset Voltage	$V_{OS}$			50			50	mV	
Offset Voltage Adjustment Range	$\pm V_{OS}$			$\pm 50$			$\pm 50$	mV	20k $\Omega$ Potentiometer
Change in Input Offset Voltage Over Temperature	$\Delta V_{OS} / \Delta T$						$\pm 5.0$	mV	$+25$ to $+85^{\circ}C$
							$\pm 5.0$	mV	$-25$ to $+25^{\circ}C$
Common Mode Rejection Ratio	CMRR	60	75		60	75		dB	$\pm 5$ volts common mode voltage
Output Voltage Swing	$\pm V_O$	$\pm 11$			$\pm 11$			V	$R_L \geq 10k\Omega$
Common Mode Voltage Range	CMVR	$\pm 10$			$\pm 10$			V	
Large Signal Voltage Gain	$A_{VOL}$	20,000	$10^5$		20,000	$10^5$		—	
Feedback Capacitance	$C_{fb}$			0.1			0.1	pF	Case guarded
Long Term Input Offset Voltage Stability	$\Delta V_{OS} / \Delta t$			$\pm 3.0$			$\pm 3.0$	mV	At $25^{\circ}C$
Slew Rate	SR		0.5		0.5			V/ $\mu s$	$R_L \geq 2k\Omega$
Input Capacitance	$C_{IN}$		0.7		0.7			pF	Case guarded
Input Capacitance	$C_{IN}$		1.5		1.5			pF	Case grounded

## CIRCUIT NOTES

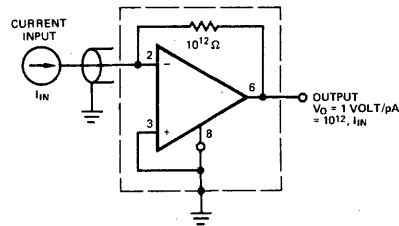
### VOLTAGE OFFSET NULL CIRCUIT



### VOLTAGE FOLLOWER



### LOW LEVEL CURRENT MEASURING CIRCUIT



NOTE: Adjust input offset voltage to  $0V \pm 10\mu V$  before measuring leakage.

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# ICH8510/8520/8530 Power Amplifier/ Motor & Actuator Driver

## KEY FEATURES:

- Delivers up to 2.7 amps @ 24-28V DC (30V supplies)
- Protected against inductive kick back with internal power limiting
- Programmable current limiting (short circuit protection)
- Package is electrically isolated (allowing easy heat sinking)
- DC gain > 100dB
- 20mA typical standby quiescent current
- Popular 8 pin TO-3 package
- Internal frequency compensation
- Can drive up to 0.1 horsepower motors.

## DESCRIPTION:

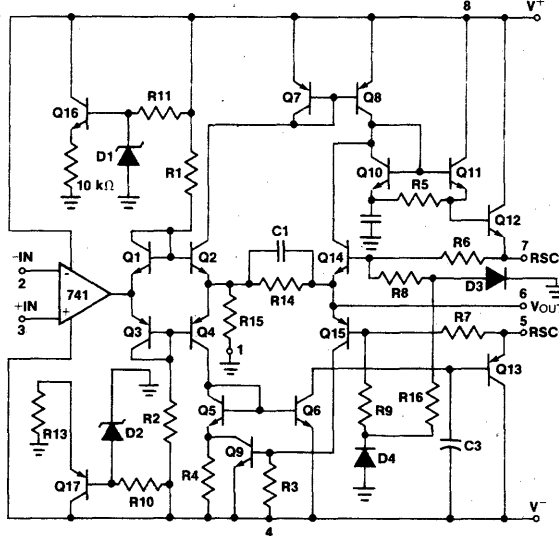
The ICH8510/8520/8530 is a family of hybrid power amplifiers that have been specifically designed to drive linear and rotary actuators, electronic valves, push-pull solenoids, and DC & AC motors.

There are three models available for up to +30V power supply operation: 2.7 amps @ 24 volt output levels, 2 amps @ 24V and 1 amp @ 24V. All amplifiers are protected against shorts to ground by the addition of 2 external protection resistors. For a device operating at lower voltages, see the ICH8515.

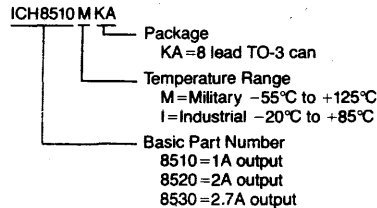
The design uses a conventional 741 operational amplifier, a special monolithic driver chip (BL8063), NPN & PNP power transistors, and internal frequency compensating capacitors. The chips are mounted on a beryllium oxide substrate for optimum heat transfer to the metal package; this substrate provides electrical isolation between amplifiers and metal package.

The I.C. power driver chip has built-in regulators to drive the 741 @ typically  $\pm 13V$  supply voltages.

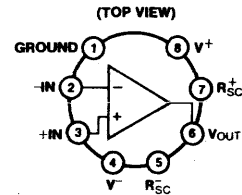
## SCHEMATIC DIAGRAM



## ORDERING INFORMATION



## PIN CONFIGURATION (outline dwg KA)



# ICH8510/8520/8530



## ABSOLUTE MAXIMUM RATINGS @ T<sub>A</sub> = 25°C

Supply Voltage .....	±32V
Power Dissipation, Safe Operating Area .....	See Curves
Differential Input Voltage .....	±30V
Input Voltage .....	±15V (Note 1)
Peak Output Current .....	See Curves (Note 2)
Output Short Circuit Duration (to ground) .....	Continuous (Note 2)
Operating Temperature Range M .....	-55°C — +125°C
I .....	-20°C — +85°C
Storage Temperature Range .....	-65°C to +150°C
Lead Temperature (Soldering, 10 seconds) .....	300°C
Max Case Temperature .....	150°C

**Note 1:** Rating applies to supply voltages of ±15V. For lower supply voltages, V<sub>INMAX</sub> = V<sub>SUPP</sub>.

**Note 2:** Ratings apply as long as package dissipation is not exceeded. Device must be mounted on heat sink, see Figures 8 and 12.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL SPECIFICATIONS T<sub>A</sub> = +25°C. V<sub>SUPP</sub> = ±30V (unless otherwise stated)

DESCRIPTION	SYMBOL	CONDITIONS	ICH8510I		ICH8510M		ICH8520I		ICH8520M		ICH8530I		ICH8530M		UNITS
			MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
Input Offset Voltage Change with Power Dissipation	ΔV <sub>OS</sub> /ΔP <sub>d</sub>	Mtd. on Wakefield 403 Heat Sink		4		2		4		2		4		2	mV/W
Input Offset Voltage	V <sub>OS</sub>	R <sub>S</sub> = 10 kΩ P <sub>d</sub> < 1W	-6	+6	-3	+3	-6	+6	-3	+3	-6	+6	-3	+3	mV
Input Bias Current	I <sub>BIAS</sub>	R <sub>S</sub> = 10 kΩ P <sub>d</sub> < 1W		500		250		500		250		500		250	nA
Input Offset Current	I <sub>OS</sub>	R <sub>S</sub> = 10 kΩ P <sub>d</sub> < 1W		200		100		200		100		200		100	nA
Large Signal Voltage Gain	A <sub>VOL</sub>	R <sub>L</sub> = 20Ω V <sub>O</sub> = 2/3 V <sub>SUPP</sub>	100		100		100		100		100		100		dB
Input Voltage Range	V <sub>CMR</sub>		-10	+10	-10	+10	-10	+10	-10	+10	-10	+10	-10	+10	V
Common Mode Rejection Ratio	CMRR	R <sub>S</sub> = 10 kΩ	70		70		70		70		70		70		dB
Power Supply Rejection Ratio	PSRR	R <sub>S</sub> = 10 kΩ	77		77		77		77		77		77		dB
Slew Rate	SR	C <sub>L</sub> = 3 pF, A <sub>V</sub> = 1 R <sub>L</sub> = 10Ω V <sub>O</sub> = 2/3 V <sub>SUPP</sub>	0.5		0.5		0.5		0.5		0.5		0.5		V/μs
Output Voltage Swing	V <sub>OMAX</sub>	R <sub>L</sub> = 20Ω A <sub>V</sub> = 10	(R <sub>L</sub> = 30Ω) ±26V		(R <sub>L</sub> = 30Ω) ±26V		±26V		±26V		±25V		±25V		V
Output Current (3)	I <sub>MAX</sub>	R <sub>L</sub> = 8Ω A <sub>V</sub> = 10	1.0		1.0		2.0		2.0		2.7		2.7		A
Power Supply Quiescent Current	I <sub>Q</sub>	R <sub>L</sub> = ∞ V <sub>IN</sub> = 0V		125		100		125		100		125		100	mA

**Note 3:** See Figure #9 if Power Supplies are less than ±30V.

## ELECTRICAL SPECIFICATIONS (continued) T<sub>A</sub> = -55°C. to +125°C.(M) or T<sub>A</sub> = -20°C. to +85°C.(I)

Input Offset Voltage	V <sub>OS</sub>	P <sub>d</sub> < 1W	-10	+10	-9	+9	-10	+10	-9	+9	-10	+10	-9	+9	MV
Input Bias Current	I <sub>BIAS</sub>	P <sub>d</sub> < 1W		1500		750		1500		750		1500		750	nA
Input Offset Current	I <sub>OS</sub>			500		200		500		200		500		200	nA
Large Signal Voltage Gain	A <sub>VOL</sub>	R <sub>L</sub> = 20Ω ΔV <sub>O</sub> = 2/3 V <sub>SUPP</sub>	90		90		90		90		90		90		dB
Output Voltage Swing	V <sub>OMAX</sub>	R <sub>L</sub> = 20Ω, A <sub>V</sub> = 10	±24		±24		±24		±24		±24		±24		V
Thermal Resistance Junction to Ambient	R <sub>θJA</sub>	Without Heat Sink		40		40		40		40		40		40	°C/W
Thermal Resistance Junction to Case	R <sub>θJC</sub>			2.5		2.5		2.5		2.5		2.5		2.5	°C/W
Thermal Resistance Junction to Ambient	R <sub>θJA</sub>	Mtd. on Wakefield 403 Heat Sink		(Typ.) 4.0		(Typ.) 4.0		(Typ.) 4.0		(Typ.) 4.0		(Typ.) 4.0		(Typ.) 4.0	°C/W
Supply Voltage Range	V <sub>SUPP</sub>		±18	±30	±18	±30	±18	±30	±18	±30	±18	±30	±18	±30	V



# ICL761X/ 762X/763X/764X Low Power CMOS™ Operational Amplifiers

## FEATURES

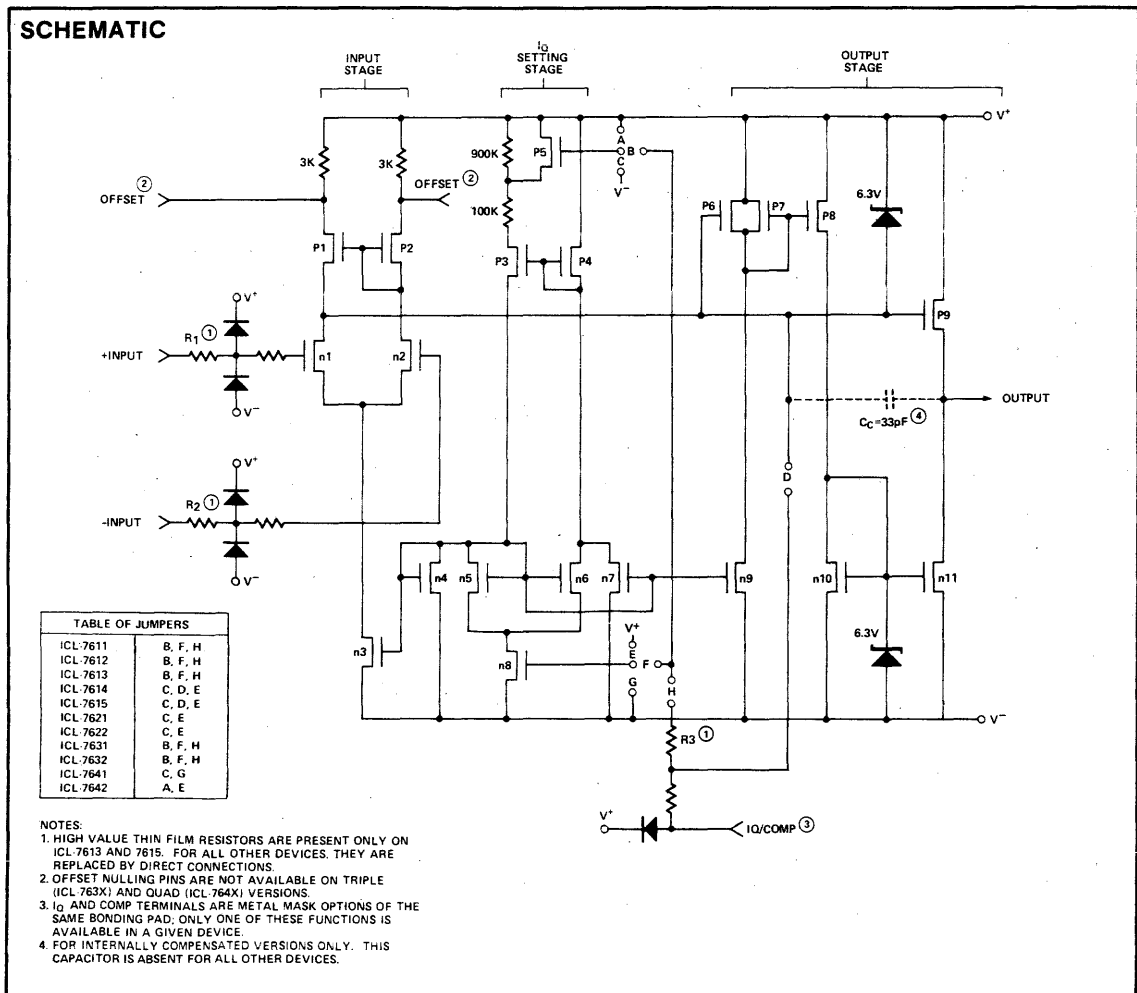
- Wide operating voltage range  $\pm 1.0V$  to  $\pm 8V$
- High input impedance —  $10^{12}\Omega$
- Programmable power consumption — as low as  $20\mu W$
- Input current lower than BIFETs — typ 1pA
- Available as singles, duals, triples, and quads
- Output voltage swing ranges to within millivolts of  $V^-$  to  $V^+$
- Low power replacement for many standard op amps
- Compensated and uncompensated versions

## APPLICATIONS

- Portable instruments
- Telephone headsets
- Hearing aid/microphone amplifiers
- Meter amplifiers
- Medical instruments
- High impedance buffers

A number of special options are available. They include:

- Single, dual, triple, and quad configurations
- Internally compensated and uncompensated versions
- Inputs protected to  $\pm 200V$  (ICL7613/15)
- Input common mode voltage range greater than supply rails (ICL7612)



# ICL761X/762X/763X/764X



## GENERAL DESCRIPTION

The ICL761X/762X/763X/764X series is a family of monolithic CMOS op amps. These amplifiers provide the designer with high performance operation at low supply voltages and selectable quiescent currents, and are an ideal design tool when ultra low input current and low power drain are essential.

The basic amplifier will operate at supply voltages ranging from  $\pm 1.0V$  to  $\pm 8V$ , and may be operated from a single Lithium cell.

A unique quiescent current programming pin allows setting of standby current to 1 mA, 100  $\mu A$ , or 10  $\mu A$ , with no external components. This results in power drain as low as 20  $\mu W$ . Output swings range to within a few millivolts of the supply voltages.

Of particular significance is the extremely low (1 pA) input current, input noise current of .01 pA/  $\sqrt{Hz}$ , and  $10^{12}\Omega$  input impedance. These features optimize performance in very high source impedance applications.

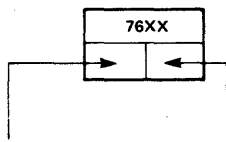
The inputs are internally protected and require no special handling procedures. Outputs are fully protected against shorts to ground or to either supply.

AC performance is excellent, with a slew rate of 1.6V/ $\mu s$ , and unity gain bandwidth of 1 MHz at  $I_Q = 1$  mA.

Because of the low power dissipation, operating temperatures and drift are quite low. Applications utilizing these features may include stable instruments, extended life designs, or high density packages.

## SELECTION GUIDE

### BASIC TYPE



OFFSET NULL CAPABILITY  
Y = YES  
N = NO.

$I_Q$  SETTING  
L = 10  $\mu A$  FIXED  
M = 100  $\mu A$  FIXED  
H = 1mA FIXED  
P = PROGRAMMABLE

### ORDERING INFORMATION <sup>[2]</sup>

ICL76XX M N O P

V<sub>OS</sub> SELECTION  
A = 2mV  
B = 5mV  
C = 10mV  
D = 15mV  
E = 20mV

TEMP. RANGE  
C = 0°C TO 70°C  
M = -55°C TO +125°C

PACKAGE CODE  
TV - TO-99, 8 PIN  
PA - PLASTIC 8 PIN MINIDIP  
PD - 14 PIN PLASTIC  
PE - 16 PIN PLASTIC  
JD - 14 PIN CERDIP  
JE - 16 PIN CERDIP

	BASIC TYPE						ORDER SUFFIX				
	COMPENSATED	EXTERNALLY COMPENSATED	COMPENSATED/ INPUT PROTECTED	EXTERNALLY COMPENSATED INPUT PROTECTED	EXTENDED CMVR						
						TO-99	MINI DIP	PLASTIC DIP [1]	CERAMIC DIP [1]	DIE	
SINGLE	7611	7614	7613	7615	7612	ACTV	AMTV	ACPA			
	Y P	Y M	Y P	Y M	Y P	BCTV	BMTV	BCPA			DC/D
						DCTV		DCPA			
DUAL 1458 PINOUT	7621					ACTV	AMTV	ACPA			
	N M					BCTV	BMTV	BCPA			DC/D
						DCTV		DCPA			
DUAL 747 PINOUT	7622							ACPD	ACJD	AMJD	
	Y M							BCPD	BCJD	BMJD	
								DCPD	DCJD		DC/D
TRIPLE	7631	7632						BCPE	BCJE	BMJE	
	N P	N P						CCPE	CCJE	CMJE	
								ECPE	ECJE		EC/D
QUAD High I <sub>Q</sub>	7641							BCPD	BCJD	BMJD	
	N H							CCPD	CCJD	CMJD	
								ECPD	ECJD		EC/D
QUAD Low I <sub>Q</sub>	7642							BCPD	BCJD	BMJD	
	N L							CCPD	CCJD	CMJD	
								ECPD	ECJD		EC/D

- NOTES: 1. Duals and quads are available in 14 pin DIP packages, triples in 16 pin only.  
2. Ordering code must consist of basic device and order suffix, e.g., ICL7611BCPA.  
3. ICL7632 is not compensatable. Recommended for use in high gain circuits only.

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# ICL761X/762X/763X/764X



## PIN CONFIGURATIONS

DEVICE	DESCRIPTION	PIN ASSIGNMENTS	
ICL7611XCPA ICL7611XCTV ICL7611XMTV ICL7612XCPA ICL7612XCTV ICL7612XMTV ICL7613XCPA ICL7613XCTV ICL7613XMTV	Internal compensation, plus offset null capability and external $I_Q$ control.	<b>TO-99 (TOP VIEW)</b> (outline dwg TV)	<b>8 PIN DIP (TOP VIEW)</b> (outline dwg PA)
ICL7614XCPA ICL7614XCTV ICL7614XMTV ICL7615XCPA ICL7615XCTV ICL7615XMTV	Fixed $I_Q$ (100 $\mu$ A), external compensation, and offset null capability.	<b>TO-99 (TOP VIEW)</b> (outline dwg TV)	<b>8 PIN DIP (TOP VIEW)</b> (outline dwg PA)
ICL7621XCPA ICL7621XCTV ICL7621XMTV	Dual op amps with internal compensation; $I_Q$ fixed at 100 $\mu$ A Pin compatible with Texas Inst. TL082, Motorola MC1458, Raytheon RC4558	<b>TO-99 (TOP VIEW)</b> (outline dwg TV)	<b>8 PIN DIP (TOP VIEW)</b> (outline dwg PA)
ICL7622XCPD	Dual op amps with internal compensation and offset null capability; $I_Q$ fixed at 100 $\mu$ A Pin compatible with Texas Inst. TL083, Fairchild $\mu$ A747	<b>14 PIN DIP (TOP VIEW)</b> (outline dwgs JD, PD)	

Note: Pins 9 and 13 are internally connected.

# ICL761X/762X/763X/764X



## PIN CONFIGURATIONS (Cont.)

DEVICE	DESCRIPTION	PIN ASSIGNMENTS
ICL7631XCPE ICL7632XCPE	Triple op amps with internal compensation (ICL7631) and no compensation (ICL7632). Adjustable $I_Q$ Same pin configuration as ICL8023.	<p><b>16 PIN DIP (TOP VIEW)</b> (outline dwgs JE, PE)</p> <p>Note: Pins 5 and 15 are internally connected.</p>
ICL7641XCPD ICL7642XCPD	Quad op amps with internal compensation. $I_Q$ fixed at 1mA (ICL7641) $I_Q$ fixed at 10 $\mu$ A (ICL7642) Pin compatible with Texas Instr. TL084 National LM324 Harris HA4741	<p><b>14 PIN DIP (TOP VIEW)</b> (outline dwg JD, PD)</p>

## GENERAL INFORMATION

### STATIC PROTECTION

All devices are static protected by the use of input diodes. However, strong static fields should be avoided, as it is possible for the strong fields to cause degraded diode junction characteristics, which may result in increased input leakage currents.

### LATCHUP AVOIDANCE

Junction-isolated CMOS circuits employ configurations which produce a parasitic 4-layer (p-n-p-n) structure. The 4-layer structure has characteristics similar to an SCR, and under certain circumstances may be triggered into a low impedance state resulting in excessive supply current. To avoid this condition, no voltage greater than 0.3V beyond the supply rails may be applied to any pin. (An exception to this rule concerns the inputs of the ICL7613 and ICL7615, which are protected to  $\pm 200V$ .) In general, the op amp supplies must be established simultaneously with, or before any input signals are applied. If this is not possible, the drive circuits must limit input current flow to 2 mA to prevent latchup.

### CHOOSING THE PROPER $I_Q$

Each device in the ICL76XX family has a similar  $I_Q$  set-up scheme, which allows the amplifier to be set to nominal quiescent currents of 10  $\mu$ A, 100  $\mu$ A or 1 mA.

These current settings change only very slightly over the entire supply voltage range. The ICL7611/12/13 and ICL7631/32 have an external  $I_Q$  control terminal, permitting user selection of each amplifier's quiescent current. (The ICL7614/15, 7621/22, and 7641/42 have fixed  $I_Q$  settings — refer to selector guide for details.) To set the  $I_Q$  of programmable versions, connect the  $I_Q$  terminal as follows:

$I_Q = 10\mu A$  —  $I_Q$  pin to  $V^+$

$I_Q = 100\mu A$  —  $I_Q$  pin to ground. If this is not possible, any voltage from  $V^+ - 0.8$  to  $V^- + 0.8$  can be used.

$I_Q = 1mA$  —  $I_Q$  pin to  $V^-$

NOTE: The negative output current available is a function of the quiescent current setting. For maximum p-p output voltage swings into low impedance loads,  $I_Q$  of 1 mA should be selected.

### OUTPUT STAGE AND LOAD DRIVING CONSIDERATIONS

Each amplifier's quiescent current flows primarily in the output stage. This is approximately 70% of the  $I_Q$  settings. This allows output swings to almost the supply rails for output loads of 1M, 100K, and 10K, using the output stage in a highly linear class A mode. In this mode, crossover distortion is avoided and the voltage gain is maximized. However, the output stage can also be operated in Class AB, which can supply

# ICL761X/762X/763X/764X



higher output currents. (See graphs under Typical Operating Characteristics). During the transition from Class A to Class B operation, the output transfer characteristic is non-linear and the voltage gain decreases.

A special feature of the output stage is that it approximates a transconductance amplifier, and its gain is directly proportional to load impedance. Approximately the same open loop gains are obtained at each of the  $I_Q$  settings if corresponding loads of 10K, 100K, and 1M are used.

## INPUT OFFSET NULLING

For those models provided with OFFSET NULLING pins, nulling may be achieved by connecting a 25K pot between the OFFSET terminals with the wiper connected to  $V^+$ . At quiescent currents of 1 mA and 100  $\mu$ A, the nulling range provided is adequate for all  $V_{OS}$  selections; however with  $I_Q = 10 \mu$ A, nulling may not be possible with higher values of  $V_{OS}$ .

## FREQUENCY COMPENSATION

The IC's 7611/12/13, 7621/22, 7631, 7641/42 are internally compensated, and are stable for closed loop gains as low as unity for capacitive loads up to 100pF

The ICL7614 and 15 are externally compensated by connecting a capacitor between the COMP and OUT pins. A 39pF capacitor is required for unity gain compensation; for greater than unity gain applications, increased bandwidth and slew rate can be obtained by reducing the value of the compensating capacitor.

Since the  $g_m$  of the first stage is proportional to  $\sqrt{I_Q}$ , greatest compensation is required when  $I_Q = 1$  mA. The ICL7632 is not compensated internally, nor can it be compensated externally. The device is stable when used as follows:

- $I_Q$  of 1 mA for gains  $\geq 20$
- $I_Q$  of 100  $\mu$ A for gains  $\geq 10$
- $I_Q$  of 10  $\mu$ A for gains  $\geq 5$

## HIGH VOLTAGE INPUT PROTECTION

The ICL7613 and 7615 include on-chip thin film resistors and clamping diodes which allow voltages of up to  $\pm 200$  to be applied to either input for an indefinite time without device failure. These devices will be useful where high common mode voltages, differential mode voltages, or high transients may be experienced. Such conditions may be found when interfacing separate systems with separate supplies. Unity gain stability is somewhat degraded with capacitive loads because of the high value of input resistors.

## EXTENDED COMMON MODE INPUT RANGE

The ICL7612 incorporates additional processing which allows the input CMVR to exceed each power supply rail by 0.1 volt for applications where  $V_{SUPP} \geq \pm 1.5V$ . For those applications where  $V_{SUPP} \leq \pm 1.5V$ , the input CMVR is limited in the positive direction, but may exceed the negative supply rail by 0.1 volt in the negative direction (eg. for  $V_{SUPP} = \pm 1.0V$ , the input CMVR would be +0.6 volts to -1.1 volts).

## OPERATION AT $V_{SUPP} = \pm 1.0$ VOLTS

Operation at  $V_{SUPP} = \pm 1.0V$  is guaranteed at  $I_Q = 10 \mu$ A only. This applies to these devices with selectable  $I_Q$ , and those devices are set internally to  $I_Q = 10 \mu$ A (i.e., ICL7611, 7612, 7613, 7631, 7632, 7642).

Output swings to within a few millivolts of the supply rails are achievable for  $R_L \geq 1$  Meg $\Omega$ . Guaranteed input CMVR is  $\pm 0.6V$  minimum and typically  $\pm 0.9V$  to  $-0.7$  at  $V_{SUPP} = \pm 1.0V$ . For applications where greater common mode range is desirable, refer to description of ICL7612 above.

The user is cautioned that, due to extremely high input impedances, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup.

## ABSOLUTE MAXIMUM RATINGS<sup>1</sup>

Total Supply Voltage $V^+$ to $V^-$ .....	18V
Input Voltage .....	$V^+ + 0.3$ to $V^- - 0.3V$
Input Voltage ICL7613/15 Only .....	$V^+ + 200$ to $V^- - 200V$
Differential Input Voltage <sup>2</sup> .....	$\pm (V^+ + 0.3 - (V^- - 0.3))V$
Differential Input Voltage <sup>2</sup> .....	$\pm (V^+ + 200 - (V^- - 200))V$
Duration of Output Short Circuit <sup>3</sup> .....	Unlimited
Continuous Power Dissipation @ 25°C .....	Above 25°C
	derate as follows:
TO-99 .....	250mW
8 Lead Minidip .....	250mW
14 Lead Plastic .....	375mW
14 Lead Cerdip .....	500mW
16 Lead Plastic .....	375mW
16 Lead Cerdip .....	500mW
Storage Temperature Range .....	-55°C to +150°C

## Operating Temperature Range

M Series .....	-55°C to +125°C
C Series .....	0°C to +70°C
Lead Temperature - Soldering, 10 sec .....	300°C

## Notes:

1. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
2. Long term offset voltage stability will be degraded if large input differential voltages are applied for long periods of time.
3. The outputs may be shorted to ground or to either supply, for  $V_{SUPP} \leq 10V$ . Care must be taken to insure that the dissipation rating is not exceeded.

# ICL761X/762X



ELECTRICAL CHARACTERISTICS  $V_{SUPP} = \pm 5.0V$ ,  $T_A = 25^\circ C$ , unless otherwise specified

PARAMETER	SYMBOL	CONDITIONS	76XXA			76XXB			76XXD			UNITS
			MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Input Offset Voltage	$V_{OS}$	$R_S \leq 100K\Omega$ , $T_A = 25^\circ C$ $T_{MIN} \leq T_A \leq T_{MAX}$			2 3			5 7			15 20	mV
Temperature Coefficient of Vos	$\Delta V_{OS}/\Delta T$	$R_S \leq 100K\Omega$		10			15			25		$\mu V/^\circ C$
Input Offset Current	$I_{OS}$	$T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$		0.5	30 300 800		0.5	30 300 800		0.5	30 300 800	pA
Input Bias Current	$I_{BIAS}$	$T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$		1.0	50 400 4000		1.0	50 400 4000		1.0	50 400 4000	pA
Common Mode Voltage Range (Except ICL7612)	$V_{CMR}$	$I_Q = 10\mu A$ <sup>1</sup> $I_Q = 100\mu A$ $I_Q = 1mA$ <sup>1</sup>	$\pm 4.4$ $\pm 4.2$ $\pm 3.7$			$\pm 4.4$ $\pm 4.2$ $\pm 3.7$			$\pm 4.4$ $\pm 4.2$ $\pm 3.7$			V
Extended Common Mode Voltage Range (ICL7612 Only)	$V_{CMP}$	$I_Q = 10\mu A$	$\pm 5.3$			$\pm 5.3$			$\pm 5.3$			V
		$I_Q = 100\mu A$	+5.3 -5.1			+5.3 -5.1			+5.3 -5.1			
		$I_Q = 1mA$	+5.3 -4.5			+5.3 -4.5			+5.3 -4.5			
Output Voltage Swing	$V_{OUT}$	(1) $I_Q = 10\mu A$ , $R_L = 1M\Omega$ $T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$	$\pm 4.9$ $\pm 4.8$ $\pm 4.7$			$\pm 4.9$ $\pm 4.8$ $\pm 4.7$			$\pm 4.9$ $\pm 4.8$ $\pm 4.7$			V
		$I_Q = 100\mu A$ , $R_L = 100k\Omega$ $T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$	$\pm 4.9$ $\pm 4.8$ $\pm 4.5$			$\pm 4.9$ $\pm 4.8$ $\pm 4.5$			$\pm 4.9$ $\pm 4.8$ $\pm 4.5$			
		(1) $I_Q = 1mA$ , $R_L = 10k\Omega$ $T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$	$\pm 4.5$ $\pm 4.3$ $\pm 4.0$			$\pm 4.5$ $\pm 4.3$ $\pm 4.0$			$\pm 4.5$ $\pm 4.3$ $\pm 4.0$			
Large Signal Voltage Gain	$A_{VOL}$	$V_O = \pm 4.0V$ , $R_L = 1M\Omega$ $I_Q = 10\mu A$ <sup>1</sup> , $T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$	86 80 74	104		80 75 68	104		80 75 68	104		dB
		$V_O = \pm 4.0V$ , $R_L = 100k\Omega$ $I_Q = 100\mu A$ , $T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$	86 80 74	102		80 75 68	102		80 75 68	102		
		$V_O = \pm 4.0V$ , $R_L = 10k\Omega$ $I_Q = 1mA$ <sup>1</sup> , $T_A = 25^\circ C$ $\Delta T_A = C$ $\Delta T_A = M$	80 76 72	83		76 72 68	83		76 72 68	83		
Unity Gain Bandwidth	$G_{BW}$	$I_Q = 10\mu A$ <sup>1</sup> $I_Q = 100\mu A$ $I_Q = 1mA$ <sup>1</sup>		0.044 0.48 1.4			0.044 0.48 1.4			0.044 0.48 1.4		MHz
Input Resistance	$R_{IN}$			10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		$\Omega$
Common Mode Rejection Ratio	$CMRR$	$R_S \leq 100K\Omega$ , $I_Q = 10\mu A$ <sup>1</sup>	76	96		70	96		70	96		dB
		$R_S \leq 100K\Omega$ , $I_Q = 100\mu A$	76	91		70	91		70	91		
		$R_S \leq 100K\Omega$ , $I_Q = 1mA$ <sup>1</sup>	66	87		60	87		60	87		
Power Supply Rejection Ratio	$PSRR$	$R_S \leq 100K\Omega$ , $I_Q = 10\mu A$ <sup>1</sup>	80	94		80	94		80	94		dB
		$R_S \leq 100K\Omega$ , $I_Q = 100\mu A$	80	86		80	86		80	86		
		$R_S \leq 100K\Omega$ , $I_Q = 1mA$ <sup>1</sup>	70	77		70	77		70	77		
Input Referred Noise Voltage	$e_n$	$R_S = 100\Omega$ , $f = 1KHz$		100			100			100		$nV/\sqrt{Hz}$
Input Referred Noise Current	$i_n$	$R_S = 100\Omega$ , $f = 1KHz$		0.01			0.01			0.01		$pA/\sqrt{Hz}$
Supply Current (Per Amplifier)	$I_{SUPP}$	No Signal, No Load $I_Q = 10\mu A$ <sup>1</sup> $I_Q = 100\mu A$ $I_Q = 1mA$ <sup>1</sup>		0.01 0.1 1.0	0.02 0.25 2.5		0.01 0.1 1.0	0.02 0.25 2.5		0.01 0.1 1.0	0.02 0.25 2.5	mA
Channel Separation	$V_{O1}/V_{O2}$	$A_{VOL} = 100$		120			120			120		dB
Slew Rate <sup>3</sup>	$SR$	$A_{VOL} = 1$ , $C_L = 100pF$ , $V_{IN} = 8V_{pp}$ $I_Q = 10\mu A$ <sup>1</sup> , $R_L = 1M\Omega$ $I_Q = 100\mu A$ , $R_L = 100K\Omega$ $I_Q = 1mA$ <sup>1</sup> , $R_L = 10K\Omega$		0.016 0.16 1.6			0.016 0.16 1.6			0.016 0.16 1.6		V/ $\mu s$
Rise Time <sup>3</sup>	$t_r$	$V_{IN} = 50mV$ , $C_L = 100pF$ $I_Q = 10\mu A$ <sup>1</sup> , $R_L = 1M\Omega$ $I_Q = 100\mu A$ , $R_L = 100K\Omega$ $I_Q = 1mA$ <sup>1</sup> , $R_L = 10K\Omega$		20 2 0.9			20 2 0.9			20 2 0.9		$\mu s$
Overshoot Factor <sup>3</sup>		$V_{IN} = 50mV$ , $C_L = 100pF$ $I_Q = 10\mu A$ <sup>1</sup> , $R_L = 1M\Omega$ $I_Q = 100\mu A$ , $R_L = 100K\Omega$ $I_Q = 1mA$ <sup>1</sup> , $R_L = 10K\Omega$		5 10 40			5 10 40			5 10 40		%

Note: 1. ICL7611, 7612, 7613 only. 2. C = Commercial Temperature Range:  $0^\circ C$  to  $+70^\circ C$   
M = Military Temperature Range:  $-55^\circ C$  to  $+125^\circ C$  3. ICL7614/15; 39pF from pin 6 to pin 8.



# ICL761X/762X



**ELECTRICAL CHARACTERISTICS**  $V_{SUPP} = \pm 1.0V$ ,  $I_Q = 10\mu A$ ,  $T_A = 25^\circ C$ , unless otherwise specified.  
Specs apply to ICL7611/7612/7613 only.

PARAMETER	SYMBOL	CONDITIONS	76XXA			76XXB			UNITS
			MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Input Offset Voltage	$V_{OS}$	$R_S \leq 100K\Omega$ , $T_A = 25^\circ C$ $T_{MIN} \leq T_A \leq T_{MAX}$			2 3			5 7	mV
Temperature Coefficient of $V_{OS}$	$\Delta V_{OS}/\Delta T$	$R_S \leq 100K\Omega$		10			15		$\mu V/^\circ C$
Input Offset Current	$I_{OS}$	$T_A = 25^\circ C$ $\Delta T_A = C$		0.5	30 300		0.5	30 300	pA
Input Bias Current	$I_{BIAS}$	$T_A = 25^\circ C$ $\Delta T_A = C$		1.0	50 500		1.0	50 500	pA
Common Mode Voltage Range (Except ICL7612)	$V_{CMR}$		$\pm 0.6$			$\pm 0.6$			V
Extended Common Mode Voltage Range (ICL7612 Only)	$V_{CMR}$		+0.6 to -1.1			+0.6 to -1.1			V
Output Voltage Swing	$V_{OUT}$	$R_L = 1M\Omega$ , $T_A = 25^\circ C$ $\Delta T_A = C$		$\pm 0.98$ $\pm 0.96$			$\pm 0.98$ $\pm 0.96$		V
Large Signal Voltage Gain	$A_{VOL}$	$V_O = \pm 0.1V$ , $R_L = 1M\Omega$ $T_A = 25^\circ C$ $\Delta T_A = C$		90 80			90 80		dB
Unity Gain Bandwidth	GBW			0.044			0.044		MHz
Input Resistance	$R_{IN}$			$10^{12}$			$10^{12}$		$\Omega$
Common Mode Rejection Ratio	CMRR	$R_S \leq 100K\Omega$		80			80		dB
Power Supply Rejection Ratio	PSRR	$R_S \leq 100K\Omega$		80			80		dB
Input Referred Noise Voltage	$e_n$	$R_S = 100\Omega$ , $f = 1KHz$		100			100		$nV/\sqrt{Hz}$
Input Referred Noise Current	$i_n$	$R_S = 100\Omega$ , $f = 1KHz$		0.01			0.01		$pA/\sqrt{Hz}$
Supply Current (Per Amplifier)	$I_{SUPP}$	No Signal, No Load		6	15		6	15	$\mu A$
Slew Rate	SR	$A_{VOL} = 1$ , $C_L = 100pF$ , $V_{IN} = 0.2V_{p-p}$ $R_L = 1M\Omega$		0.016			0.016		$V/\mu s$
Rise Time	$t_r$	$V_{IN} = 50mV$ , $C_L = 100pF$ $R_L = 1M\Omega$		20			20		$\mu s$
Overshoot Factor		$V_{IN} = 50mV$ , $C_L = 100pF$ $R_L = 1M\Omega$		5			5		%

Note: C = Commercial Temperature Range ( $0^\circ C$  to  $+70^\circ C$ ); M = Military Temperature Range ( $-55^\circ C$  to  $+125^\circ C$ ).

# ICL7650 Chopper Stabilized Operational Amplifier

## FEATURES

- Extremely low input offset voltage –  $1\mu\text{V}$  over temperature range
- Low long-term and temperature drifts of input offset voltage
- Low DC input bias current –  $10\text{pA}$
- Extremely high gain, CMRR and PSRR – min 120dB
- High slew rate –  $2.5\text{V}/\mu\text{s}$
- Wide bandwidth – 2MHz
- Internally compensated for unity-gain operation
- Very low intermodulation effects (open loop phase shift  $< 10^\circ$  @ chopper frequency)
- Clamp circuit to avoid overload recovery problems and allow comparator use
- Extremely low chopping spikes at input and output

## GENERAL DESCRIPTION

The ICL7650 chopper-stabilized amplifier is a high-performance device which offers exceptionally low offset voltage and input-bias parameters, combined with excellent bandwidth and speed characteristics. Intersil's unique CMOS approach to chopper-stabilized amplifier design yields a versatile precision component which can replace more expensive hybrid or modular parts, while at the same time out-performing them and other monolithic devices.

The chopper amplifier achieves its low offset by comparing the inverting and non-inverting input voltages in a nulling amplifier, nulled by alternate clock phases. Two external capacitors are required to store the correcting potentials on the two amplifier nulling inputs; these are the only external components necessary.

The clock oscillator and all the other control circuitry is entirely self-contained, however the 14-pin version includes a provision for the use of an external clock, if required for a particular application. In addition, the ICL7650 is internally compensated for unity-gain operation.

## ORDERING INFORMATION

PART	TEMP RANGE	PACKAGE
ICL 7650 CPA	$0^\circ - 70^\circ\text{C}$	8-Pin Plastic
ICL 7650 CPD	$0^\circ - 70^\circ\text{C}$	14-Pin Plastic
ICL 7650 CTV	$0^\circ - 70^\circ\text{C}$	8-Pin TO-99
ICL 7650 IJA	$-20^\circ\text{C} - 85^\circ\text{C}$	8-Pin Cerdip
ICL 7650 IJD	$-20^\circ\text{C} - 85^\circ\text{C}$	14-Pin Cerdip
ICL 7650 ITV	$-20^\circ\text{C} - 85^\circ\text{C}$	8-Pin TO-99
ICL 7650 MJD	$-55^\circ\text{C} - 125^\circ\text{C}$	14-Pin Cerdip
ICL 7650 MTV	$-55^\circ\text{C} - 125^\circ\text{C}$	8-Pin TO-99
ICL 7650 CPA-1	$0^\circ - 70^\circ\text{C}$	8-Pin Plastic
ICL 7650 CTV-1	$0^\circ - 70^\circ\text{C}$	8-Pin TO-99
ICL 7650 IJD-1	$-20^\circ\text{C} - 85^\circ\text{C}$	8-Pin Cerdip
ICL 7650 ITV-1	$-20^\circ\text{C} - 85^\circ\text{C}$	8-Pin TO-99
ICL 7650 MTV-1	$-55^\circ\text{C} - 125^\circ\text{C}$	8-Pin TO-99

NOTE: By using the ICL 7650-1 versions and connecting C<sub>RETN</sub>, better noise performance can be attained.

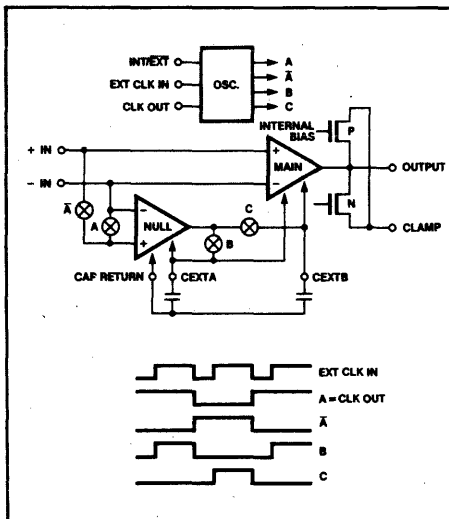
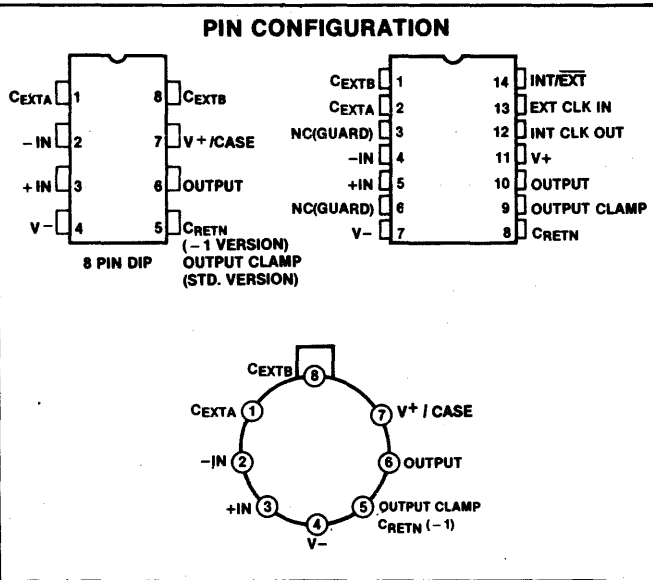
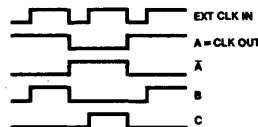


FIG. 1 BLOCK DIAGRAM



## PIN CONFIGURATION

Intersil LINEAR

# ICL7650



## ABSOLUTE MAXIMUM RATINGS

Total Supply Voltage ( $V^+$  to  $V^-$ ) ..... 18 Volts  
 Input Voltage ..... ( $V^+ + 0.3$ ) to ( $V^- - 0.3$ ) Volts  
 Storage Temp. Range .....  $-55^\circ\text{C}$  to  $150^\circ\text{C}$   
 Operating Temp. Range ..... See Note 1  
 Lead Temperature (Soldering, 10 sec) .....  $300^\circ\text{C}$   
 Voltage on oscillator control pins .....  $V^+$  to  $V^-$   
 except EXT CLOCK IN: ( $V^+ + 0.3$ ) to ( $V^+ - 6.0$ ) Volts  
 Duration of Output short circuit ..... Indefinite  
 Current into any pin .....  $10\text{mA}$   
 — while operating (Note 4) .....  $100\ \mu\text{A}$

Cont. Total Power Dissipn ( $T_A = 25^\circ\text{C}$ )  
 CERDIP Package ..... 500 mW  
 Plastic Package ..... 375 mW  
 TO-99 ..... 250 mW

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**OPERATING CHARACTERISTICS:** Test Conditions:  $V^+ = +5\text{V}$ ,  $V^- = -5\text{V}$ ,  $T_A = +25^\circ\text{C}$ , Test Ckt (unless otherwise specified)

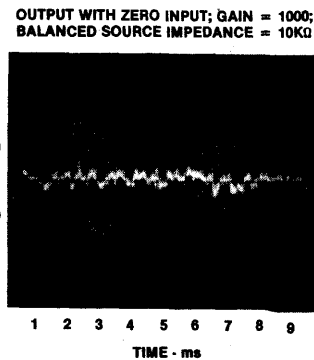
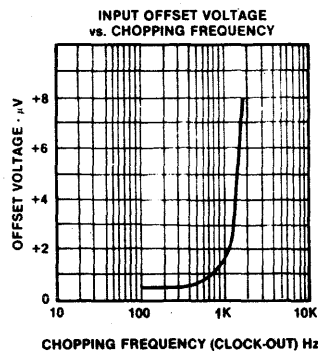
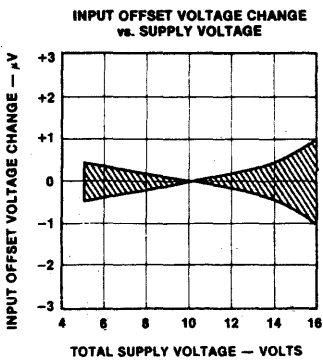
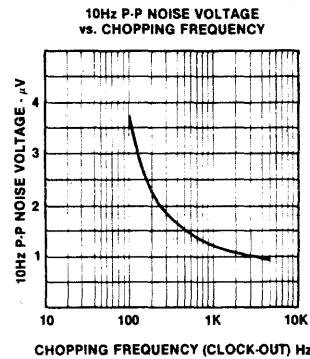
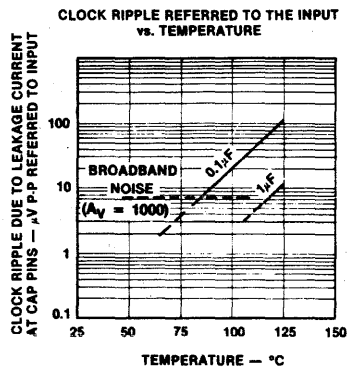
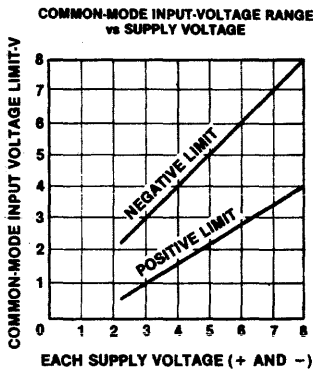
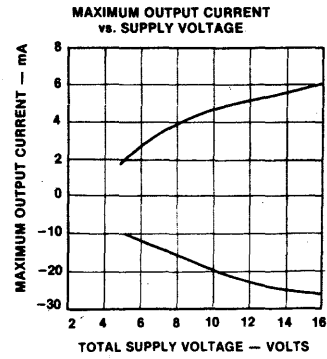
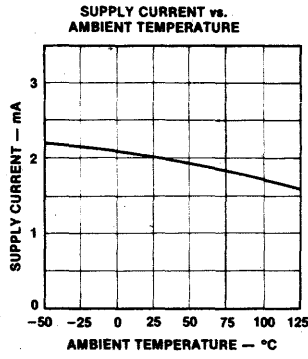
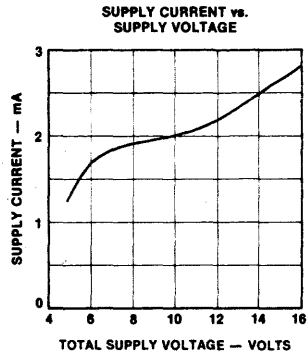
PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	LIMITS TYP.	MAX.	UNIT
Input Offset Voltage	$V_{OS}$	$T_A = +25^\circ\text{C}$ $-55^\circ\text{C} < T_A < +85^\circ\text{C}$ $-55^\circ\text{C} < T_A < +125^\circ\text{C}$		$\pm 0.7$ $\pm 1.0$	$\pm 5$ 5.0	$\mu\text{V}$
Average Temp. Coefficient of Input Offset Voltage	$\frac{\Delta V_{OS}}{\Delta T}$	$-20^\circ\text{C} < T_A < +85^\circ\text{C}$		0.01 50	0.05	$\mu\text{V}/^\circ\text{C}$
Input Bias Current (doubles every $10^\circ\text{C}$ )	$I_{BIAS}$	$T_A = +25^\circ\text{C}$ $0^\circ\text{C} < T_A < +70^\circ\text{C}$ $-20^\circ\text{C} < T_A < +85^\circ\text{C}$		1.5 35 100	10	pA
Input Offset Current	$I_{OS}$	$T_A = 25^\circ\text{C}$		0.5		pA
Input Resistance	$R_{IN}$			$10^{12}$		$\Omega$
Large Signal Voltage Gain	$A_{VOL}$	$R_L = 10\text{k}\Omega$	$1 \times 10^6$	$5 \times 10^6$		V/V
Output Voltage Swing (Note 3)	$V_{OUT}$	$R_L = 10\text{k}\Omega$ $R_L = 100\text{k}\Omega$	$\pm 4.7$	$\pm 4.85$ $\pm 4.95$		V
Common Mode Voltage Range	CMVR		-5.0	-5.2 to +2.0	1.6	V
Common Mode Rejection Ratio	CMRR	CMVR = $-5\text{V}$ to $+1.6$	120	130		dB
Power Supply Rejection Ratio	PSRR	$\pm 3\text{V}$ to $\pm 8\text{V}$	120	130		dB
Input Noise Voltage	$e_{npp}$	$R_S = 100\Omega$ 0 to 10Hz		2		$\mu\text{Vp-p}$
Input Noise Current	$i_n$	$f = 10\text{Hz}$		0.01		$\text{pA}/\sqrt{\text{Hz}}$
Unity Gain Bandwidth	GBW			2.0		MHz
Slew Rate	SR	$C_L = 50\text{pF}$ , $R_L = 10\text{k}\Omega$		2.5		$\text{V}/\mu\text{s}$
Rise Time	$t_r$			0.2		$\mu\text{s}$
Overshoot				20		%
Operating Supply Range	$V^+$ to $V^-$		4.5		16	V
Supply Current	$I_{SUPP}$	no load		2.0	3.5	mA
Internal Chopping Frequency	$f_{ch}$	pins 12-14 open (DIP)	120	200	375	Hz
Clamp ON Current (note 2)		$R_L = 100\text{k}\Omega$	25	70	200	$\mu\text{A}$
Clamp OFF Current (note 2)		$-4.0\text{V} < V_{OUT} < +4.0\text{V}$		1		pA
Offset Voltage vs Time				100		$\text{nV}/\sqrt{\text{month}}$

NOTE 1: Operating temperature range for M series parts is  $-55^\circ\text{C}$  to  $+125^\circ\text{C}$ , for I series is  $-20^\circ\text{C}$  to  $+85^\circ\text{C}$ , for C series is  $0^\circ\text{C}$  to  $+70^\circ\text{C}$   
 NOTE 2: See OUTPUT CLAMP under detailed description.  
 NOTE 3: OUTPUT CLAMP not connected. See typical characteristic curves for output swing vs clamp current characteristics.  
 NOTE 4: Limiting input current to  $100\ \mu\text{A}$  is recommended to avoid latch-up problems. Typically  $1\text{mA}$  is safe, however this is not guaranteed.  
 NOTE 5:  $I_{OS} = 2 \cdot I_{BIAS}$

LINEAR

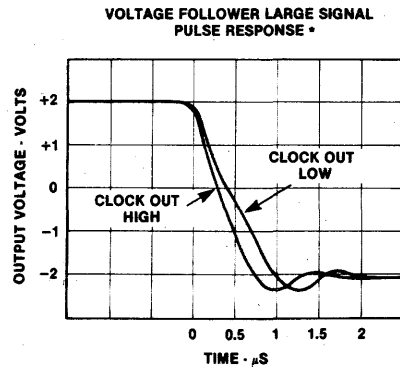
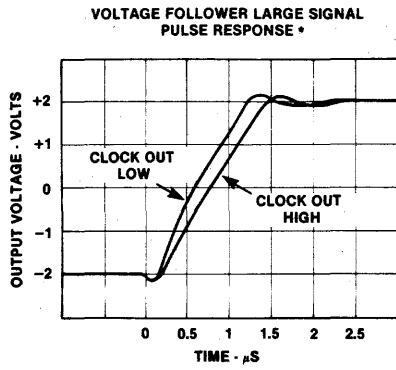
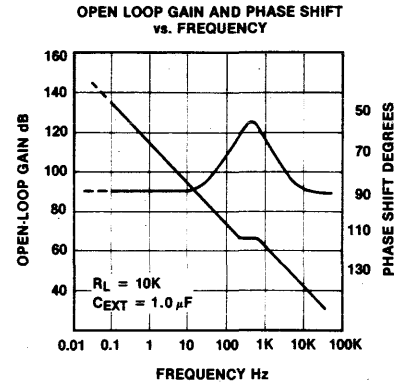
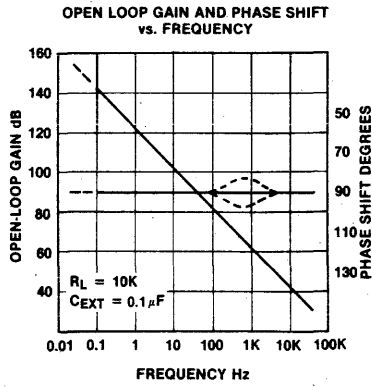
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TYPICAL OPERATING CHARACTERISTICS

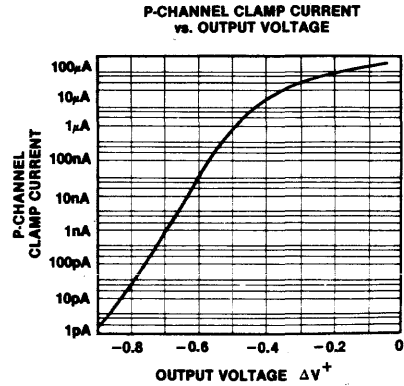
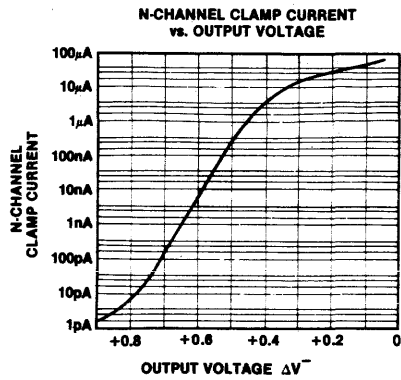


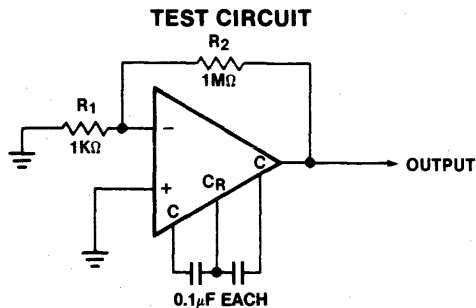
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\* THE TWO DIFFERENT RESPONSES CORRESPOND TO THE TWO PHASES OF THE CLOCK.





## DETAILED DESCRIPTION

### AMPLIFIER

The block diagram shows the major elements of the ICL7650. There are two amplifiers, the main amplifier, and the nulling amplifier; both have offset-null capability. The main amplifier is connected full-time from the input to the output, while the nulling amplifier, under the control of the chopping frequency oscillator and clock circuit, alternately nulls itself and the main amplifier. The nulling connections, which are MOSFET back gates, are inherently high impedance, and two external capacitors provide the required storage of the nulling potentials and the necessary nulling-loop time constants. The nulling arrangement operates over the full common-mode and power-supply ranges, and is also independent of the output level, thus giving exceptionally high CMRR, PSRR, and  $A_{VOL}$ .

Careful balancing of the input switches, and the inherent balance of the input circuit, minimizes chopper frequency charge injection at the input terminals, and also the feedforward-type injection into the compensation capacitor, which is the main cause of output spikes in this type of circuit.

### INTERMODULATION

Previous chopper-stabilized amplifiers have suffered from intermodulation effects between the chopper frequency and input signals. These arise because the finite AC gain of the amplifier necessitates a small AC signal at the input. This is seen by the zeroing circuit as an error signal, which is chopped and fed back, thus injecting sum and difference frequencies and causing disturbances to the gain and phase vs. frequency characteristics near the chopping frequency. These effects are substantially reduced in the ICL7650 by feeding the nulling circuit with a dynamic current, corresponding to the compensation capacitor current, in such a way as to cancel that portion of the input signal due to finite AC gain. Since that is the major error contribution to the ICL7650, the intermodulation and gain/phase disturbances are held to very low values, and can generally be ignored.

### CAPACITOR CONNECTION

The null-storage capacitors should be connected to the  $C_{EXTA}$  and  $C_{EXTB}$  pins, with a common connection to the  $C_{RETN}$  pin (in the case of 14-pin devices) or the  $V^-$  pin (in the case of the 8-pin devices). This connection should be made directly by either a separate wire or PC trace to avoid injecting load current IR drops into

the capacitive circuitry. The outside foil, where available, should be connected to  $C_{RETN}$  (or  $V^-$ ).

### OUTPUT CLAMP

The OUTPUT CLAMP pin allows reduction of the overload recovery time inherent with chopper-stabilized amplifiers. When tied to the inverting input pin, or summing junction, a current path between this point and the OUTPUT pin occurs just before the device output saturates. Thus uncontrolled input differential inputs are avoided, together with the consequent charge build-up on the correction-storage capacitors. The output swing is slightly reduced.

### CLOCK

The ICL7650 has an internal oscillator giving a chopping frequency of 200 Hz, available at the CLOCK OUT pin on the 14-pin devices. Provision has also been made for the use of an external clock in these parts. The INT/EXT pin has an internal pull-up and may be left open for normal operation, but to utilize an external clock this pin must be tied to  $V^-$  to disable the internal clock. The external clock signal may then be applied to the EXT. CLOCK IN pin. At low frequencies, the duty cycle of the external clock is not critical, since an internal divide-by-two provides the desired 50% switching duty cycle. However, since the capacitors are charged only when EXT CLK IN is HIGH, a 50-80% positive duty cycle is favored for frequencies above 500Hz to ensure that any transients have time to settle before the capacitors are turned OFF. The external clock should swing between  $V^+$  and GROUND for power supplies up to  $\pm 6V$ , and between  $V^+$  and  $V^+ - 6V$  for higher supply voltages. Note that a signal of about 400Hz will be present at the EXT CLK IN pin with INT/EXT high or open. This is the internal clock signal before the divider.

In those applications where a strobe signal is available, an alternate approach to avoid capacitor misbalancing during overload can be used. If a strobe signal is connected to EXT CLK IN so that it is low during the time that the overload signal is applied to the amplifier, neither capacitor will be charged. Since the leakage at the capacitor pins is quite low at room temperature, the typical amplifier will drift less than  $10\mu V/sec$ , and relatively long measurements can be made with little change in offset.

## BRIEF APPLICATION NOTES

### COMPONENT SELECTION

The two required capacitors,  $C_{EXTA}$  and  $C_{EXTB}$ , have optimum values depending on the clock or chopping frequency. For the preset internal clock, the correct value is  $0.1\mu F$ , and to maintain the same relationship between the chopping frequency and the nulling time constant this value should be scaled approximately in proportion if an external clock is used. A high-quality film-type capacitor such as mylar is preferred, although a ceramic or other lower-grade capacitor may prove suitable in many applications. For quickest settling on initial turn-on, low dielectric absorption capacitors (such as polypropylene) should be used. With ceramic capacitors, several seconds may be required to settle to  $1\mu V$ .

**STATIC PROTECTION**

All device pins are static-protected by the use of input diodes. However, strong static fields and discharges should be avoided, as they can cause degraded diode junction characteristics, which may result in increased input-leakage currents.

**LATCH-UP AVOIDANCE**

Junction-isolated CMOS circuits inherently include a parasitic 4-layer (p-n-p-n) structure which has characteristics similar to an SCR. Under certain circumstances this junction may be triggered into a low-impedance state, resulting in excessive supply current. To avoid this condition, no voltage greater than 0.3V beyond the supply rails should be applied to any pin. In general, the amplifier supplies must be established either at the same time or before any input signals are applied. If this is not possible, the drive circuits must limit input current flow to under 1mA to avoid latchup, even under fault conditions.

**OUTPUT STAGE/LOAD DRIVING**

The output circuit is a high-impedance stage (approximately 18kΩ), and therefore, with loads less than this the chopper amplifier behaves in some ways like a transconductance amplifier whose open-loop gain is proportional to load resistance. For example, the open-loop gain will be 17dB lower with a 1kΩ load than with a 10kΩ load. If the amplifier is used strictly for DC, this lower gain is of little consequence, since the DC gain is typically greater than 120dB even with a 1KΩ load. However, for wideband applications, the best frequency response will be achieved with a load resistor of 10K or higher. This will result in a smooth 6dB/octave response from 0.1Hz to 2MHz, with phase shifts of less than 10° in the transition region where the main amplifier takes over from the null amplifier.

**THERMO-ELECTRIC EFFECTS**

The ultimate limitations to ultra-high precision DC amplifiers are the thermo-electric or Peltier effects arising in thermocouple junctions of dissimilar metals, alloys, silicon, etc. Unless all junctions are at the same

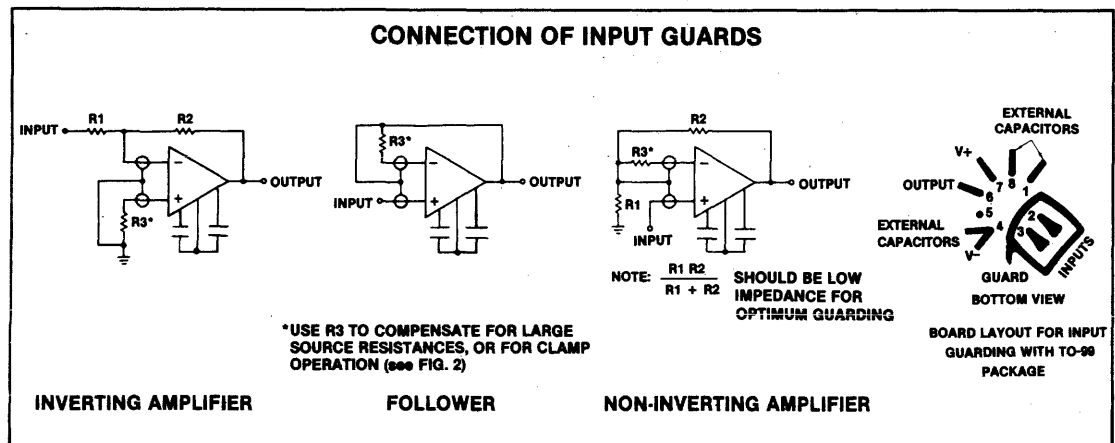
temperature, thermoelectric voltages typically around 0.1μV/°C, but up to tens of μV/°C for some materials, will be generated. In order to realize the extremely low offset voltages that the chopper amplifier can provide, it is essential to take special precautions to avoid temperature gradients. All components should be enclosed to eliminate air movement, especially that caused by power-dissipating elements in the system. Low thermoelectric-coefficient connections should be used where possible and power supply voltages and power dissipation should be kept to a minimum. High-impedance loads are preferable, and good separation from surrounding heat-dissipating elements is advisable.

**GUARDING**

Extra care must be taken in the assembly of printed circuit boards to take full advantage of the low input currents of the ICL7650. Boards must be thoroughly cleaned with TCE or alcohol and blown dry with compressed air. After cleaning, the boards should be coated with epoxy or silicone rubber to prevent contamination.

Even with properly cleaned and coated boards, leakage currents may cause trouble, particularly since the input pins are adjacent to pins that are at supply potentials. This leakage can be significantly reduced by using guarding to lower the voltage difference between the inputs and adjacent metal runs. Input guarding of the 8-lead TO-99 package is accomplished by using a 10-lead pin circle, with the leads of the device formed so that the holes adjacent to the inputs are empty when it is inserted in the board. The guard, which is a conductive ring surrounding the inputs, is connected to a low impedance point that is at approximately the same voltage as the inputs. Leakage currents from high-voltage pins are then absorbed by the guard.

The pin configuration of the 14-pin dual in-line package is designed to facilitate guarding, since the pins adjacent to the inputs are not used (this is different from the standard 741 and 101A pin configuration, but corresponds to that of the LM108).



Intersil LINEAR

# ICL7650



## PIN COMPATIBILITY

The basic pinout of the 8-pin device corresponds, where possible, to that of the industry-standard 8-pin devices, the LM741, LM101, etc. The null-storing external capacitors are connected to pins 1 and 8, usually used for offset null or compensation capacitors, or simply not connected. The output-clamp pin (5) is similarly used. In the case of the OP-05 and OP-07 devices, the replacement of the offset-null pot, connected between pins 1 and 8 and  $V+$ , by two capacitors from those pins to  $V-$ , will provide easy compatibility. As for the LM108, replacement the compensation capacitor between pins 1 and 8 by the two capacitors to  $V-$  is all that is necessary. The same operation, with the removal of any connection to pin 5, will suffice for the LM101,  $\mu A748$ , and similar parts.

The 14-pin device pinout corresponds most closely to that of the LM108 device, owing to the provision of "NC" pins for guarding between the input and all other pins. Since this device does not use any of the extra pins, and has no provision for offset-nulling, but requires a compensation capacitor, some changes will be required in layout to convert to the ICL7650.

## TYPICAL APPLICATIONS

Clearly the applications of the ICL7650 will mirror those of other op. amps. Thus, anywhere that the performance of a circuit can be significantly improved by a reduction of input-offset voltage and bias current, the ICL7650 is the logical choice. Basic non-inverting and inverting amplifier circuits are shown in Figs. 2 and 3. Both circuits can use the output clamping circuit to enhance the overload recovery performance. The only limitations on the replacement of other op. amps by

the ICL7650 are the supply voltage ( $\pm 8V$  max.) and the output drive capability ( $10k\Omega$  load for full swing). Even these limitations can be overcome using a simple booster circuit, as shown in Fig. 4, to enable the full output capabilities of the LM741 (or any other standard device) to be combined with the input capabilities of the ICL7650. The pair form a composite device, so loop gain stability, when the feedback network is added, should be watched carefully.

Fig. 5 shows the use of the clamp circuit to advantage in a zero-offset comparator. The usual problems in using a chopper stabilized amplifier in this application are avoided, since the clamp circuit forces the inverting input to follow the input signal. The threshold input must tolerate the output clamp current  $\approx V_{IN}/R$  without disturbing other portions of the system.

Normal logarithmic amplifiers are limited in dynamic range in the voltage-input mode by their input-offset voltage. The built-in temperature compensation and convenience features of the ICL8048 can be extended to a voltage-input dynamic range of close to 6 decades by using the ICL7650 to offset-null the ICL8048, as shown in Fig. 6. The same concept can also be used with such devices as the HA2500 or HA2600 families of op. amps. to add very low offset voltage capability to their very high slew rates and bandwidths. Note that these circuits will also have their DC gains, CMRR, and PSRR enhanced.

Mixing the ICL7650 with circuits operating at  $\pm 15V$  supplies requires the provision of a lower voltage. Although this can be met fairly easily, a highly efficient voltage divider can be built using the ICL7660 voltage converter circuit 'backwards'. A suitable connection is shown in Fig. 7.

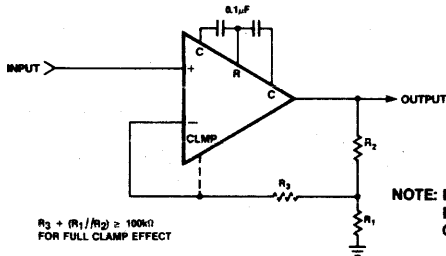


FIG. 2 NON INVERTING AMPLIFIER WITH (OPTIONAL) CLAMP

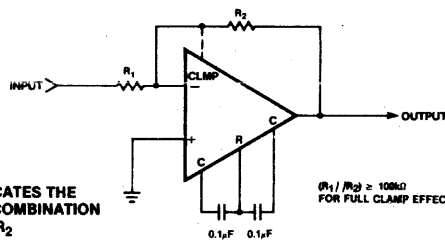


FIG. 3 INVERTING AMPLIFIER WITH (OPTIONAL) CLAMP

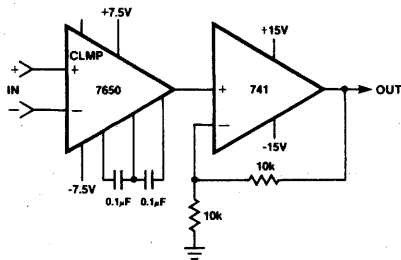


FIG. 4 USING 741 TO BOOST OUTPUT DRIVE CAPABILITY

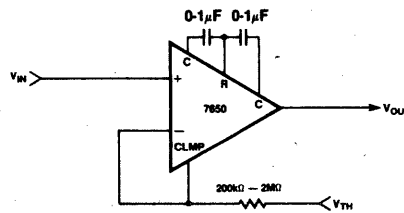


FIG. 5 LOW OFFSET COMPARATOR

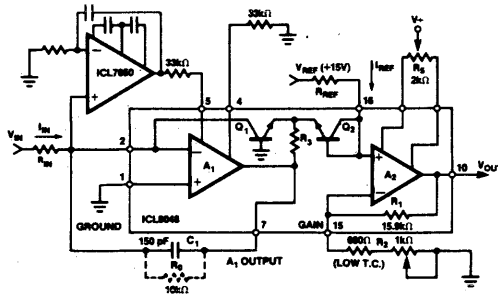
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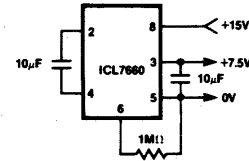
**ICL7650**



**TYPICAL APPLICATIONS (Continued)**

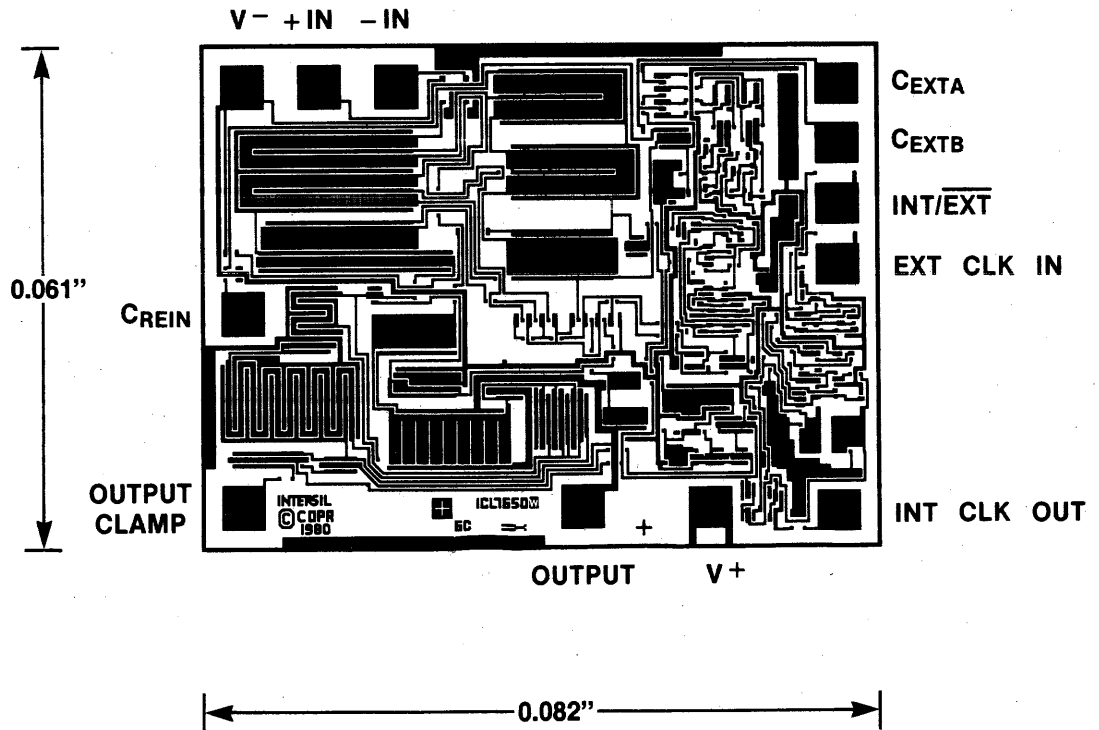


**FIG. 6 ICL8048 OFFSET NULLED BY ICL7650**



**FIG. 7 SPLITTING +15V WITH ICL7660. SAME FOR -15V. >95% EFF.**

FOR FURTHER APPLICATIONS ASSISTANCE, SEE A053 AND R017



PRELIMINARY  
Some Specifications Subject to Change Without Notice

# ICL7652

## Chopper-Stabilized Operational Amplifier

### FEATURES

- Extremely low input offset voltage— $1\mu\text{V}$  over temperature range
- Ultra low long-term and temperature drifts of input offset voltage ( $100\text{nV}/\sqrt{\text{month}}$ ,  $10\text{nV}/^\circ\text{C}$ )
- Low DC input bias current— $15\text{pA}$
- Extremely high gain, CMRR and PSRR—min 110dB
- Low input noise voltage— $0.2\mu\text{Vp-p}$  (DC—1Hz)
- Internally compensated for unity-gain operation
- Very low intermodulation effects (open-loop phase shift  $< 2^\circ$  @ chopper frequency)
- Clamp circuit to avoid overload recovery problems and allow comparator use
- Extremely low chopping spikes at input and output

### GENERAL DESCRIPTION

The ICL7652 chopper-stabilized amplifier offers exceptionally low input offset voltage and is extremely stable with

respect to time and temperature. It is similar to INTERSIL's ICL7650 but offers improved noise performance and a wider common-mode input voltage range. The bandwidth and slew rate are reduced slightly.

INTERISL's unique CMOS chopper-stabilized amplifier circuitry is user-transparent, virtually eliminating the traditional chopper amplifier problems of intermodulation effects, chopping spikes, and overrange lock-up.

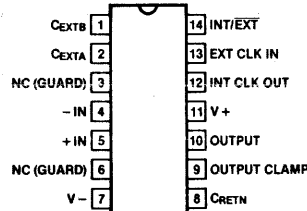
The chopper amplifier achieves its low offset by comparing the inverting and non-inverting input voltages in a nulling amplifier, nulled by alternate clock phases. Two external capacitors are required to store the correcting potentials on the two amplifier nulling inputs; these are the only external components necessary.

The clock oscillator and all the other control circuitry is entirely self-contained, however the 14-pin version includes a provision for the use of an external clock, if required for a particular application. In addition, the ICL7652 is internally compensated for unity-gain operation.

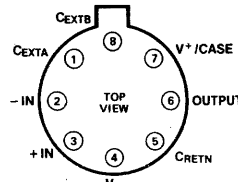
### ORDERING INFORMATION

TEMP RANGE	PACKAGE	ORDER#
0°C to +70°C	14-pin plastic	ICL7652CPD
-20°C to +85°C	14-pin CERDIP	ICL7652IJD
0°C to +70°C	8-pin TO-99	ICL7652CTV
-20°C to +85°C	8-pin TO-99	ICL7652ITV

### PIN CONFIGURATIONS



14 Lead



TO-99

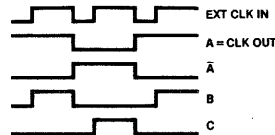
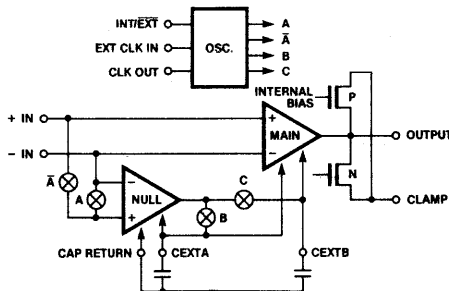


Figure 1. Block Diagram

**ICL7652**



**ABSOLUTE MAXIMUM RATINGS**

Total Supply Voltage ( $V^+$ to $V^-$ )	18V
Input Voltage	( $V^+ + 0.3$ ) to ( $V^- - 0.3$ ) V
Storage Temperature Range	-55°C to 150°C
Operating Temperature Range	See Note 1
Lead Temperature (Soldering, 10 sec)	300°C
Voltage on Oscillator Control Pins	$V^+$ to $V^-$
Duration of Output Short Circuit	Indefinite

Current into Any Pin	10mA
—while operating (Note 4)	100 $\mu$ A
Continuous Total Power Dissipation ( $T_A = 25^\circ\text{C}$ )	
CERDIP Package	500mW
Plastic Package	375mW
TO-99	250mW

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**OPERATING CHARACTERISTICS:** Test Conditions:  $V^+ = +5\text{V}$ ,  $V^- = -5\text{V}$ ,  $T_A = +25^\circ\text{C}$ , Test Circuit (unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
Input Offset Voltage	$V_{OS}$	$T_A = +25^\circ\text{C}$		$\pm 0.7$	$\pm 5$	$\mu\text{V}$
		Over Operating Temperature Range (Note 1)		$\pm 1.0$		
Average Temperature Coefficient of Input Offset Voltage	$\frac{\Delta V_{OS}}{\Delta T}$	Operating Temperature Range (Note 1)		0.01	0.05	$\mu\text{V}/^\circ\text{C}$
Input Bias Current (Doubles every 10°C above about 60°C)	$I_{BIAS}$	$T_A = +25^\circ\text{C}$		15	30	pA
		$0^\circ\text{C} < T_A < +70^\circ\text{C}$		35		
		$-20^\circ\text{C} < T_A < +85^\circ\text{C}$		100		
Input Offset Current	$I_{OS}$	$T_A = +25^\circ\text{C}$		25	60	pA
Input Resistance	$R_{IN}$			$10^{12}$		$\Omega$
Large Signal Voltage Gain	$A_{VOL}$	$R_L = 10\text{k}\Omega$ , $V_{OUT} = \pm 4\text{V}$	120	150		dB
Output Voltage Swing (Note 3)	$V_{OUT}$	$R_L = 10\text{k}\Omega$	$\pm 4.7$	$\pm 4.85$		V
		$R_L = 100\text{k}\Omega$		$\pm 4.95$		
Common-Mode Voltage Range	CMVR		-4.3	-4.8 to +4.0	3.5	V
Common-Mode Rejection Ratio	CMRR	CMVR = -4.3V to +3.5V	110	130		dB
Power Supply Rejection Ratio	PSRR	$\pm 3\text{V}$ to $\pm 8\text{V}$	110	130		dB
Input Noise Voltage	$e_{npp}$	$R_S = 100\Omega$ , DC to 1Hz		0.2		$\mu\text{Vp-p}$
		DC to 10Hz		0.7		
Input Noise Current	$i_n$	$f = 10\text{Hz}$		0.01		$\text{pA}/\sqrt{\text{Hz}}$
Unity-Gain Bandwidth	GBW			0.45		MHz
Slew Rate	SR	$C_L = 50\text{pF}$ , $R_L = 10\text{k}\Omega$		0.5		$\text{V}/\mu\text{s}$
Rise Time	$t_r$			0.8		$\mu\text{s}$
Overshoot				20		%
Operating Supply Range	$V^+$ to $V^-$		5.0		16	V
Supply Current	$I_{SUPP}$	No Load		2.0	3.5	mA
Internal Chopping Frequency	$f_{ch}$	Pins 12-14 Open (DIP)		400		Hz
Clamp ON Current (Note 2)		$R_L = 100\text{k}\Omega$	25	100		$\mu\text{A}$
Clamp OFF Current (Note 2)		$-4.0\text{V} < V_{OUT} < +4.0\text{V}$		1		pA
Offset Voltage vs Time				100		$\text{nV}/\sqrt{\text{month}}$

**Note 1:** Operating temperature range for I series parts is -20°C to +85°C, for C series is 0°C to +70°C.

**Note 2:** See OUTPUT CLAMP under detailed description.

**Note 3:** OUTPUT CLAMP not connected. See typical characteristics curves for output swing vs clamp current characteristics.

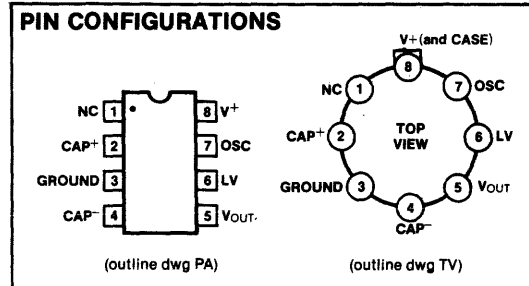
**Note 4:** Limiting input current to 100 $\mu$ A is recommended to avoid latch-up problems. Typically 1mA is safe, however this is not guaranteed.

### FEATURES

- Simple Conversion of +5V Logic Supply to ±5V Supplies
- Simple Voltage Multiplication ( $V_{OUT} = (-) nV_{IN}$ )
- 99.9% Typical Open Circuit Voltage Conversion Efficiency
- 98% Typical Power Efficiency
- Wide Operating Voltage Range 1.5V to 10.0V
- Easy to use - Requires only 2 External Non-Critical Passive Components

### APPLICATIONS

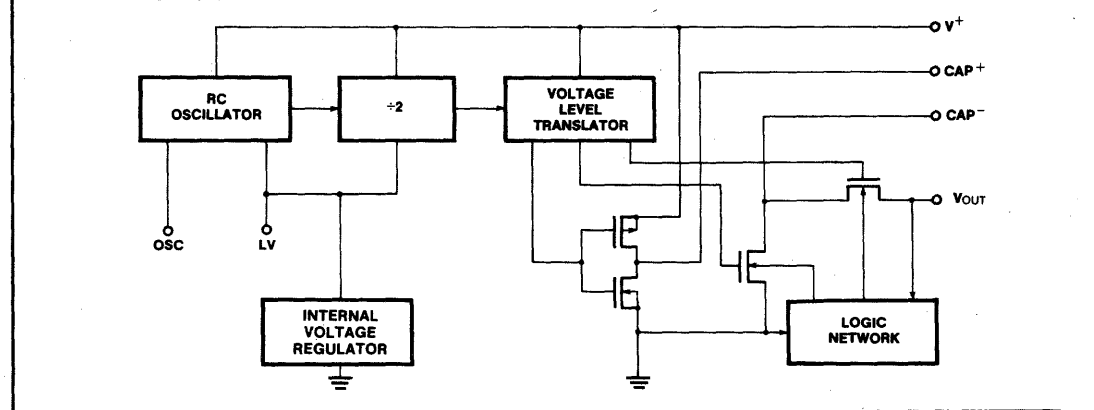
- On Board Negative Supply for up to 64 Dynamic RAMs.
- Localized  $\mu$ -Processor (8080 type) Negative Supplies
- Inexpensive Negative Supplies
- Data Acquisition Systems



### ORDERING INFORMATION

PART NUMBER	TEMP. RANGE	PACKAGE
ICL7660CTV	-20° to +70° C	TO-99
ICL7660CPA	-20° to +70° C	8 PIN MINI DIP
ICL7660MTV	-55° to +125° C	TO-99
ICL7660/D		DICE

### BLOCK DIAGRAM



### GENERAL DESCRIPTION

The Intersil ICL7660 is a monolithic MAXCMOS™ power supply circuit which offers unique performance advantages over previously available devices. The ICL7660 performs the complete supply voltage conversion from positive to negative for an input range of +1.5V to +10.0V, resulting in complementary output voltages of -1.5 to -10.0V with the addition of only 2 non-critical external capacitors needed for the charge pump and charge reservoir functions. Note that an additional diode is required for  $V_{SUPPLY} > 6.5V$ .

Contained on chip are a series DC power supply regulator, RC oscillator, voltage level translator, four output power MOS switches, and a unique logic element which senses the most negative voltage in the device and ensures that the output N-channel switches are not forward biased. This assures latch-up free operation.

The oscillator, when unloaded, oscillates at a nominal frequency of 10kHz for an input supply voltage of 5.0 volts. This frequency can be lowered by the addition of an external capacitor to the "OSC" terminal, or the oscillator may be overdriven by an external clock.

The "LV" terminal may be tied to GROUND to bypass the internal series regulator and improve low voltage (LV) operation. At medium to high voltages (+3.5 to +10.0 volts), the LV pin is left floating to prevent device latchup.

Typical applications for the ICL7660 will be data acquisition and microprocessor based systems where there is a +5 volt supply available for the digital functions and an additional -5 volt supply is required for the analog functions. The ICL7660 is also ideally suited for providing low current, -5V body bias supply for dynamic RAMs.

# ICL7660



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage	10.5V	Operating Temperature Range	
LV and OSC Input Voltage		ICL7660M	-55°C to +125°C
(Note 1)	-0.3V to (V <sup>+</sup> +0.3V) for V <sup>+</sup> < 5.5V (V <sup>+</sup> -5.5V) to (V <sup>+</sup> +0.3V) for V <sup>+</sup> > 5.5V	ICL7660C	0°C to 70°C
Current into LV (Note 1)	20μA for V <sup>+</sup> > 3.5V	Storage Temperature Range	-65°C to 150°C
Output Short Duration (V <sub>SUPPLY</sub> ≤ 5.5V)	Continuous	Lead Temperature	
Power Dissipation (Note 2)		(Soldering, 10 sec.)	300°C
ICL7660CTV	500mW		
ICL7660CPA	300mW		
ICL7660MTV	500mW		

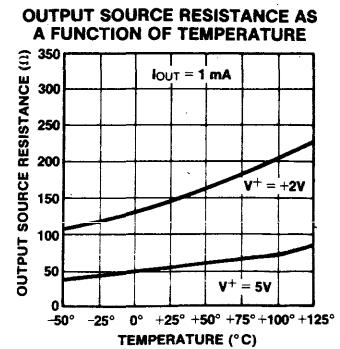
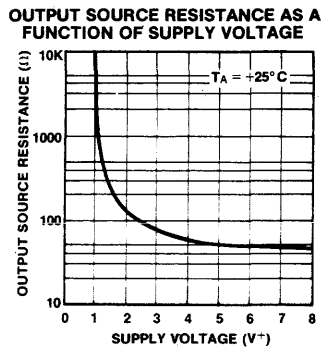
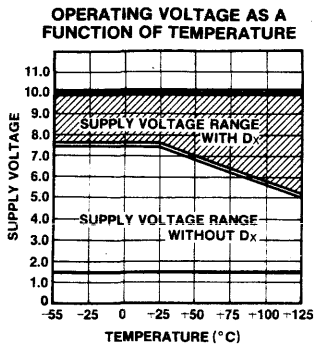
Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## OPERATING CHARACTERISTICS V<sup>+</sup> = 5V, T<sub>A</sub> = 25°C, C<sub>OSC</sub> = 0, Test Circuit Figure 1 (unless otherwise specified)

SYMBOL	PARAMETER	LIMITS			UNIT	TEST CONDITIONS
		MIN.	TYP.	MAX.		
I <sup>+</sup>	Supply Current		170	500	μA	R <sub>L</sub> = ∞
V <sup>+</sup> H1	Supply Voltage Range - Hi (D <sub>X</sub> out of circuit) (Note 3)	3.0		6.5	V	0°C ≤ T <sub>A</sub> ≤ 70°C, R <sub>L</sub> = 10kΩ, LV Open
V <sup>+</sup> L1	Supply Voltage Range - Lo (D <sub>X</sub> out of circuit)	1.5		3.5	V	-55°C ≤ T <sub>A</sub> ≤ 125°C, R <sub>L</sub> = 10kΩ, LV Open
V <sup>+</sup> H2	Supply Voltage Range - Hi (D <sub>X</sub> in circuit)	3.0		10.0	V	MIN ≤ T <sub>A</sub> ≤ MAX, R <sub>L</sub> = 10kΩ, LV Open
V <sup>+</sup> L2	Supply Voltage Range - Lo (D <sub>X</sub> in circuit)	1.5		3.5	V	MIN ≤ T <sub>A</sub> ≤ MAX, R <sub>L</sub> = 10kΩ, LV to GROUND
R <sub>OUT</sub>	Output Source Resistance		55	100	Ω	I <sub>OUT</sub> = 20mA, T <sub>A</sub> = 25°C
				120	Ω	I <sub>OUT</sub> = 20mA, -20°C ≤ T <sub>A</sub> ≤ +70°C
				150	Ω	I <sub>OUT</sub> = 20mA, -55°C ≤ T <sub>A</sub> ≤ +125°C (Note 3)
				300	Ω	V <sup>+</sup> = 2V, I <sub>OUT</sub> = 3mA, LV to GROUND -20°C ≤ T <sub>A</sub> ≤ +70°C
				400	Ω	V <sup>+</sup> = 2V, I <sub>OUT</sub> = 3mA, LV to GROUND, -55°C ≤ T <sub>A</sub> ≤ +125°C, D <sub>X</sub> in circuit (Note 3)
f <sub>OSC</sub>	Oscillator Frequency		10		kHz	
P <sub>Ef</sub>	Power Efficiency	95	98		%	R <sub>L</sub> = 5kΩ
V <sub>OUT Ef</sub>	Voltage Conversion Efficiency	97	99.9		%	R <sub>L</sub> = ∞
Z <sub>OSC</sub>	Oscillator Impedance		1.0		MΩ	V <sup>+</sup> = 2 Volts
			100		kΩ	V = 5 Volts

- Notes:**
1. Connecting any input terminal to voltages greater than V<sup>+</sup> or less than GROUND may cause destructive latchup. It is recommended that no inputs from sources operating from external supplies be applied prior to "power up" of the ICL7660.
  2. Derate linearly above 50°C by 5.5mW/°C.
  3. ICL7660M only.

## TYPICAL PERFORMANCE CHARACTERISTICS (Circuit of Figure 1)



**TYPICAL APPLICATIONS**

**1. Simple Negative Voltage Converter**

The majority of applications will undoubtedly utilize the ICL7660 for generation of negative supply voltages. Figure 4 shows typical connections to provide a negative supply where a positive supply is available. A similar scheme may be employed for supply voltages anywhere in the operating range of +1.5V to +10.0 volts, keeping in mind that pin 6 (LV) is tied to the supply negative (GND) only for supply voltages below 3.5 volts, and that diode D<sub>x</sub> must be included for proper operation at higher voltages and/or elevated temperatures.

The output characteristics of the circuit in Figure 4 are those of a nearly ideal voltage source in series with 70 ohms. Thus for a load current of -10mA and a supply voltage of +5 volts, the output voltage will be -4.3 volts. The dynamic output impedance due to the capacitor impedances is approximately  $1/\omega C$  where

$$C = C_1 = C_2$$

$$\text{giving } \frac{1}{\omega C} = \frac{1}{2\pi f_{osc} \times 10^{-5}} = 3 \text{ ohms}$$

for  $C = 10\mu\text{F}$  and  $f_{osc} = 5\text{kHz}$  (1/2 of oscillator frequency)

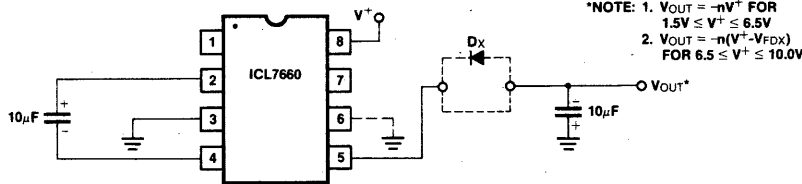


Figure 4: Simple Negative Converter

**2. Paralleling Devices**

Any number of ICL7660 voltage converters may be paralleled to reduce output resistance. The reservoir capacitor, C<sub>2</sub>, serves all devices while each device requires

its own pump capacitor, C<sub>1</sub>. The resultant output resistance would be approximately

$$R_{OUT} = \frac{R_{OUT} \text{ (of ICL7660)}}{n \text{ (number of devices)}}$$

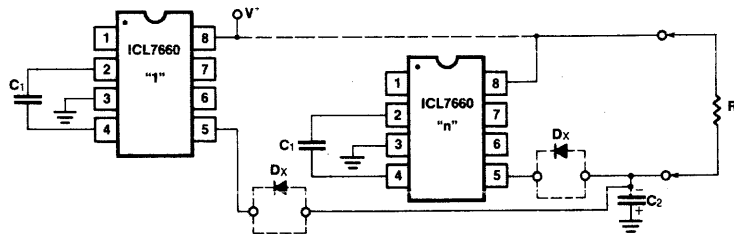


Figure 5: Paralleling Devices

**3. Cascading Devices**

The ICL7660 may be cascaded as shown to produce larger negative multiplication of the initial supply voltage, however, due to the finite efficiency of each device, the practical limit is 10 devices for light loads. The output voltage is

defined by:

$$V_{OUT} = -n(V_{IN}),$$

where n is an integer representing the number of devices cascaded. The resulting output resistance would be approximately the weighted sum of the individual ICL7660 R<sub>OUT</sub> values.

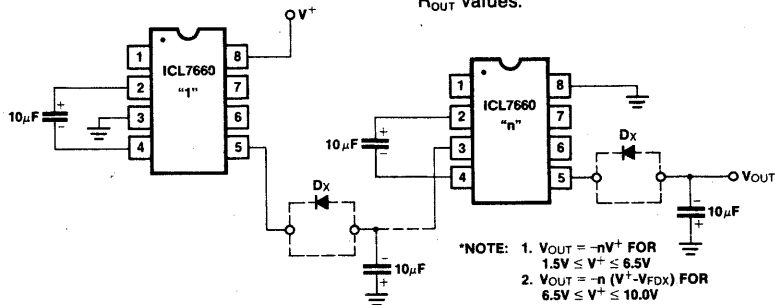
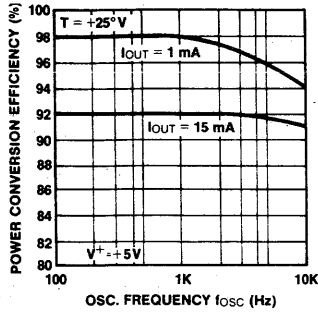


Figure 6: Cascading Devices for Increased Output Voltage

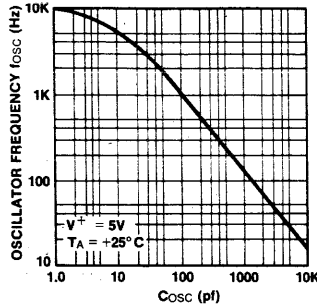
# ICL7660



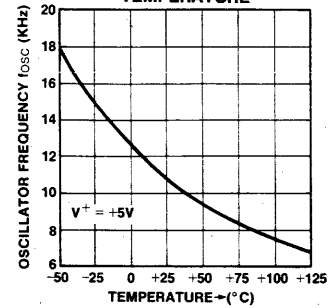
**POWER CONVERSION EFFICIENCY AS A FUNCTION OF OSC. FREQUENCY**



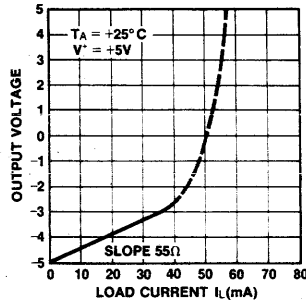
**FREQUENCY OF OSCILLATION AS A FUNCTION OF EXTERNAL OSC. CAPACITANCE**



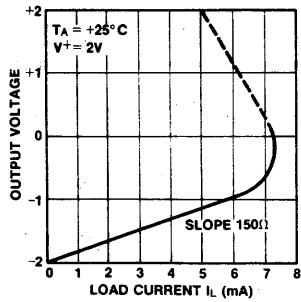
**UNLOADED OSCILLATOR FREQUENCY AS A FUNCTION OF TEMPERATURE**



**OUTPUT VOLTAGE AS A FUNCTION OF OUTPUT CURRENT**



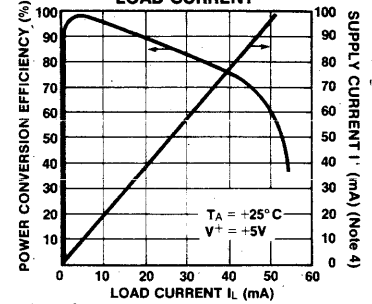
**OUTPUT VOLTAGE AS A FUNCTION OF OUTPUT CURRENT**



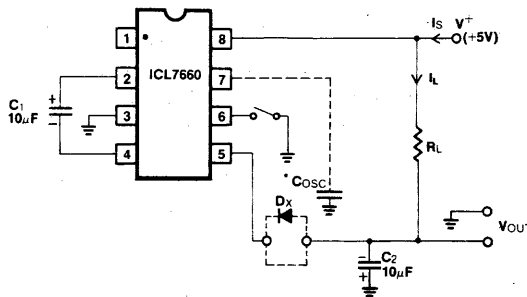
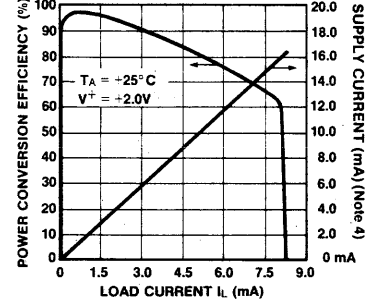
**NOTE 4.**

Note that the curves on the right include in the supply current that current fed directly into the load  $R_L$  from  $V^+$  - see Figure 1. Thus, approximately half the supply current goes directly to the positive side of the load, and the other half, through the ICL7660, to the negative side of the load. Ideally,  $V_{out} \approx 2 V_{in}$ ,  $I_s \approx 2 I_L$ , so  $V_{in} \cdot I_s = V_{out} \cdot I_L$ .

**SUPPLY CURRENT & POWER CONVERSION EFFICIENCY AS A FUNCTION OF LOAD CURRENT**



**SUPPLY CURRENT POWER CONVERSION EFFICIENCY AS A FUNCTION OF LOAD CURRENT**



- NOTES:**
1. For large value of  $C_{osc}$  ( $>1000\text{pF}$ ) the values of  $C_1$  and  $C_2$  should be increased to  $100\mu\text{F}$ .
  2.  $D_x$  is required for supply voltages greater than  $6.5\text{V}$  @  $-55^\circ \leq T_A \leq +70^\circ\text{C}$ ; refer to performance curves for additional information.

Figure 1: ICL7660 Test Circuit

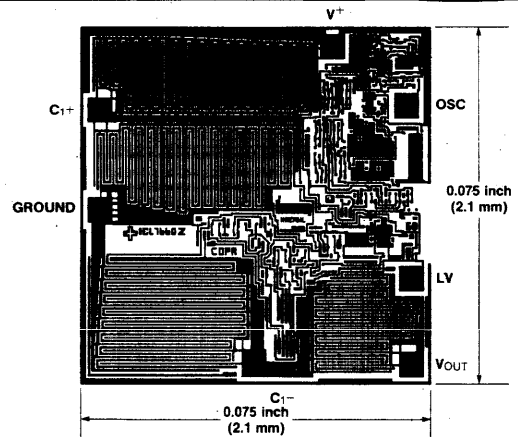


Figure 2: Chip Topography

LINEAR

Intersil

## CIRCUIT DESCRIPTION

The ICL7660 contains all the necessary circuitry to complete a voltage doubler, with the exception of 2 external capacitors which may be inexpensive 10 $\mu$ F polarized electrolytic capacitors. The mode of operation of the device may be best understood by considering Figure 3, which shows an idealized voltage doubler. Capacitor C<sub>1</sub> is charged to a voltage, V<sup>+</sup>, for the half cycle when switches S<sub>1</sub> and S<sub>3</sub> are closed. (Note: Switches S<sub>2</sub> and S<sub>4</sub> are open during this half cycle.) During the second half cycle of operation, switches S<sub>2</sub> and S<sub>4</sub> are closed, with S<sub>1</sub> and S<sub>3</sub> open, thereby shifting capacitor C<sub>1</sub> negatively by V<sup>+</sup> volts. Charge is then transferred from C<sub>1</sub> to C<sub>2</sub> such that the voltage on C<sub>2</sub> is exactly V<sup>+</sup>, assuming ideal switches and no load on C<sub>2</sub>. The ICL7660 approaches this ideal situation more closely than existing non-mechanical circuits.

In the ICL7660, the 4 switches in Figure 3 are MOS power switches; S<sub>1</sub> is a P-channel device and S<sub>2</sub>, S<sub>3</sub> & S<sub>4</sub> are N-channel devices. The main difficulty with this approach is that in integrating the switches, the substrates of S<sub>3</sub> & S<sub>4</sub> must always remain reverse biased with respect to their sources, but not so much as to degrade their "ON" resistances. In addition, at circuit startup, and under output short circuit conditions (V<sub>OUT</sub> = V<sup>+</sup>), the output voltage must be sensed and the substrate bias adjusted accordingly. Failure to accomplish this would result in high power losses and probable device latchup.

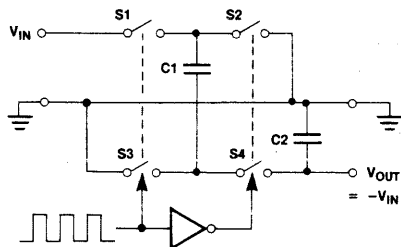


Figure 3. Idealized Voltage Doubler

This problem is eliminated in the ICL7660 by a logic network which senses the output voltage (V<sub>OUT</sub>) together with the level translators and switches the substrates of S<sub>3</sub> & S<sub>4</sub> to the correct level to maintain necessary reverse bias.

The voltage regulator portion of the ICL7660 is an integral part of the anti-latchup circuitry, however its inherent voltage drop can degrade operation at low voltages. Therefore, to improve low voltage operation the "LV" pin should be connected to GROUND, disabling the regulator. For supply voltages greater than 3.5 volts the LV terminal must be left open to insure latchup proof operation, and prevent device damage.

## THEORETICAL POWER EFFICIENCY CONSIDERATIONS

In theory a voltage multiplier can approach 100% efficiency if certain conditions are met:

- A The drive circuitry consumes minimal power
- B The output switches have extremely low ON resistance and virtually no offset.
- C The impedances of the pump and reservoir capacitors are negligible at the pump frequency.

The ICL7660 approaches these conditions for negative voltage multiplication if large values of C<sub>1</sub> and C<sub>2</sub> are used. **ENERGY IS LOST ONLY IN THE TRANSFER OF CHARGE BETWEEN CAPACITORS IF A CHANGE IN VOLTAGE OCCURS.** The energy lost is defined by:

$$E = 1/2 C_1 (V_1^2 - V_2^2)$$

where V<sub>1</sub> and V<sub>2</sub> are the voltages on C<sub>1</sub> during the pump and transfer cycles. If the impedances of C<sub>1</sub> and C<sub>2</sub> are relatively high at the pump frequency (refer to Fig. 3) compared to the value of R<sub>L</sub>, there will be a substantial difference in the voltages V<sub>1</sub> and V<sub>2</sub>. Therefore it is not only desirable to make C<sub>2</sub> as large as possible to eliminate output voltage ripple, but also to employ a correspondingly large value for C<sub>1</sub> in order to achieve maximum efficiency of operation.

## DO'S AND DON'TS

- 1 Do not exceed maximum supply voltages.
- 2 Do not connect LV terminal to GROUND for supply voltages greater than 3.5 volts.
- 3 Do not short circuit the output to V<sup>+</sup> supply for supply voltages above 5.5 volts for extended periods, however, transient conditions including startup are okay.
- 4 When using polarized capacitors, the + terminal of C<sub>1</sub> must be connected to pin 2 of the ICL7660 and the + terminal of C<sub>2</sub> must be connected to GROUND.
- 5 Add diode D<sub>x</sub> as shown in Fig. 1 for hi-voltage, elevated temperature applications.

## CONSIDERATIONS FOR HI VOLTAGE &amp; ELEVATED TEMPERATURE

The ICL7660 will operate efficiently over its specified temperature range with only 2 external passive components (storage & pump capacitors), provided the operating supply voltage does not exceed 6.5 volts at +70°C and 5.0 volts at +125°C. Exceeding these maximums at the temperatures indicated may result in destructive latch-up of the ICL7660. (Ref: Graph "Operating Voltage Vs. Temperature")

Operation at supply voltages of up to 10.0 volts over the full temperature range without danger of latch-up can be achieved by adding a general purpose diode in series with the ICL7660 output, as shown by "D<sub>x</sub>" in the circuit diagrams. The effect of this diode on overall circuit performance is the reduction of output voltage by one diode drop (approximately 0.6 volts).



## ICL7660



### 4. Changing the ICL7660 Oscillator Frequency

It may be desirable in some applications, due to noise or other considerations, to increase the oscillator frequency. This is achieved by overdriving the oscillator from an external clock, as shown in Figure 7. In order to prevent possible device latchup, a 1kΩ resistor must be used in series with the clock output. In the situation where the designer has generated the external clock frequency using TTL logic, the addition of a 10kΩ pullup resistor to V<sup>-</sup> supply is required. Note that the pump frequency with external clocking, as with internal clocking, will be 1/2 of the clock frequency. Output transitions occur on the positive-going edge of the clock.

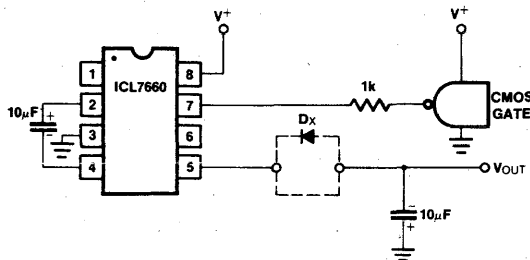


Figure 7: External Clocking

It is also possible to increase the conversion efficiency of the ICL7660 at low load levels by lowering the oscillator frequency. This reduces the switching losses, and is achieved by connecting an additional capacitor, C<sub>OSC</sub>, as shown in Figure 8. However, lowering the oscillator frequency will cause an undesirable increase in the impedance of the pump (C<sub>1</sub>) and reservoir (C<sub>2</sub>) capacitors; this is overcome by increasing the values of C<sub>1</sub> and C<sub>2</sub> by the same factor that the frequency has been reduced. For example, the addition of a 100pF capacitor between pin 7 (Osc) and V<sup>+</sup> will lower the oscillator frequency to 1kHz from its nominal frequency of 10kHz (a multiple of 10), and thereby necessitate a corresponding increase in the value of C<sub>1</sub> and C<sub>2</sub> (from 10µF to 100µF).

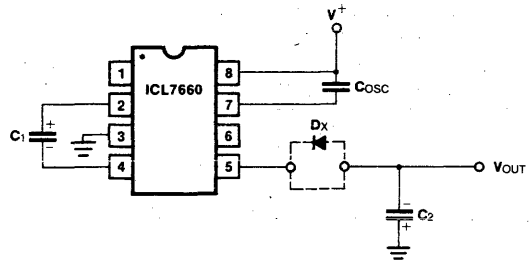


Figure 8: Lowering Oscillator Frequency

### 5. Positive Voltage Multiplication

The ICL7660 may be employed to achieve positive voltage multiplication using the circuit shown in Figure 9. In this application, the pump inverter switches of the ICL7660 are used to charge C<sub>1</sub> to a voltage level of V<sup>+</sup> - V<sub>F</sub> (where V<sup>+</sup> is the supply voltage and V<sub>F</sub> is the forward voltage drop of diode D<sub>1</sub>). On the transfer cycle, the voltage on C<sub>1</sub> plus the supply voltage (V<sup>+</sup>) is applied through diode D<sub>2</sub> to capacitor C<sub>2</sub>. The voltage thus created on C<sub>2</sub> becomes (2V<sup>+</sup>) - (2V<sub>F</sub>) or twice the supply voltage minus the combined forward voltage drops of diodes D<sub>1</sub> and D<sub>2</sub>.

The source impedance of the output (V<sub>OUT</sub>) will depend on the output current, but for V<sup>+</sup> = 5 volts and an output current of 10mA it will be approximately 60 ohms.

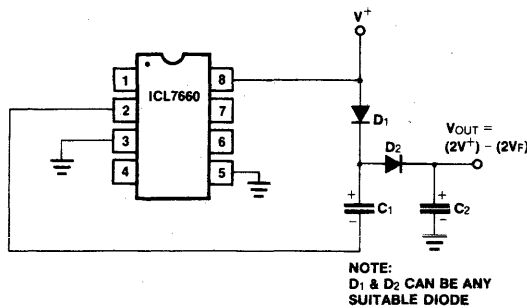


Figure 9: Positive Voltage Multiplier

### 6. Combined Negative Voltage Conversion and Positive Supply Multiplication

Figure 10 combines the functions shown in Figures 4 and 9 to provide negative voltage conversion and positive voltage multiplication simultaneously. This approach would be, for example, suitable for generating +9 volts and -5 volts from an existing +5 volt supply. In this instance capacitors C<sub>1</sub> and C<sub>3</sub> perform the pump and reservoir functions respectively for the generation of the negative voltage, while capacitors C<sub>2</sub> and C<sub>4</sub> are pump and reservoir respectively for the multiplied positive voltage. There is a penalty in this configuration which combines both functions, however, in that the source impedances of the generated supplies will be somewhat higher due to the finite impedance of the common charge pump driver at pin 2 of the device.

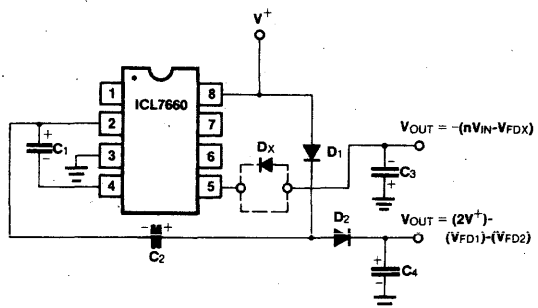


Figure 10: Combined Negative Converter and Positive Multiplier

# ICL7660



## 7. Voltage Splitting

The bidirectional characteristics can also be used to split a higher supply in half, as shown in Figure 11. The combined load will be evenly shared between the two sides. Once again, a high value resistor to the LV pin ensures start-up. Because the switches share the load in parallel, the output impedance is much lower than in the standard circuits, and higher currents can be drawn from the device. By using this circuit, and then the circuit of Figure 6, +15V can be converted (via +7.5, and -7.5) to a nominal -15V, though with rather high series resistance (~250Ω).

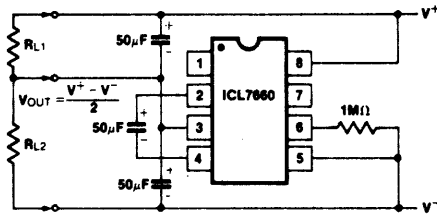


Figure 11: Splitting a Supply in Half.

ICL7660's output does not respond instantaneously to a change in input, but only after the switching delay. The circuit shown supplies enough delay to accommodate the 7660, while maintaining adequate feedback. An increase in pump and storage capacitors is desirable, and the values shown provides an output impedance of less than 5Ω to a load of 10mA.

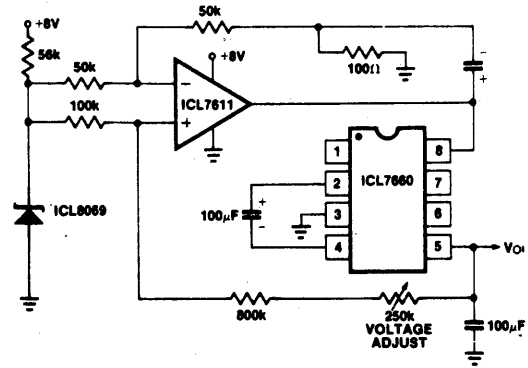


Figure 12: Regulating the Output Voltage

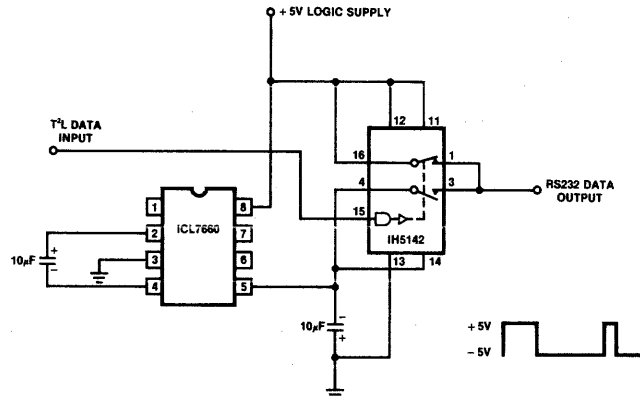


Figure 13: RS232 Levels from a Single 5V Supply

## 8. Regulated Negative Voltage Supply

In some cases, the output impedance of the ICL7660 can be a problem, particularly if the load current varies substantially. The circuit of Figure 12 can be used to overcome this by controlling the input voltage, via an ICL7611 low-power CMOS op amp, in such a way as to maintain a nearly constant output voltage. Direct feedback is inadvisable, since the

## OTHER APPLICATIONS

Further information on the operation and use of the ICL7660 may be found in A051 "Principals and Applications of the ICL7660 CMOS Voltage Converter" by Peter Bradshaw and Dave Bingham.



PRELIMINARY  
Specifications Subject To Change Without Notice

# ICL7663/7664

## CMOS Programmable Micropower Voltage Regulators

### FEATURES

- Ideal for battery-operated systems: less than 4 $\mu$ A typical current drain
- Will handle input voltages from 1.6V to 16V
- Very low input-output differential voltage
- 1.3V bandgap voltage reference
- Up to 40mA output current
- Output shutdown via current-limit sensing or external logic signal
- Output voltages programmable from 1.3V to 16V
- Output voltages with programmable negative temperature coefficients (ICL7663 only)

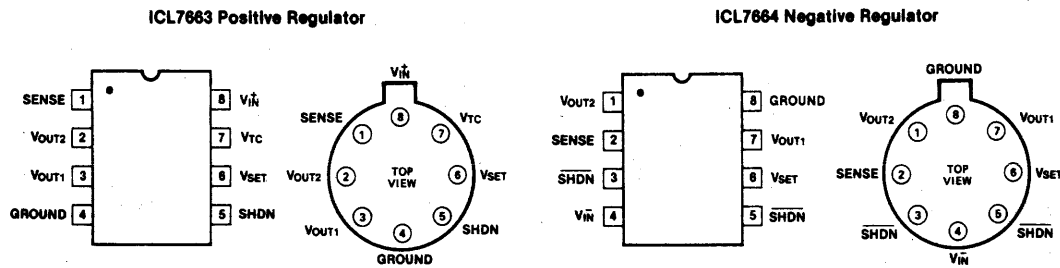
### GENERAL DESCRIPTION

The ICL7663 (positive) and ICL7664 (negative) series regulators are low-power, high-efficiency devices which accept inputs from 1.6V to 16V and provide adjustable outputs over the same range at currents up to 40mA. Operating current is typically less than 4 $\mu$ A, regardless of load.

Output current sensing and remote shutdown are available on both devices, thereby providing protection for the regulators and the circuits they power. A unique feature, on the ICL7663 only, is a negative temperature coefficient output. This can be used, for example, to efficiently tailor the voltage applied to a multiplexed LCD through the driver (e.g., ICM7231/2/3/4) so as to extend the display operating temperature range many times.

The ICL7663 and ICL7664 are available in either an 8-pin plastic minidip package or a TO-99 can.

### PIN CONFIGURATIONS (outline dwgs PA, TV)



### ORDERING INFORMATION

Positive Regulator		
ICL7663CPA	0°C to +70°C	8-pin minidip
ICL7663CTV	0°C to +70°C	TO-99
ICL7663/D		DICE

Negative Regulator		
ICL7664CPA	0°C to +70°C	8-pin minidip
ICL7664CTV	0°C to +70°C	TO-99
ICL7664/D		DICE

LINEAR

Intersil

**ABSOLUTE MAXIMUM RATINGS, ICL7663 POSITIVE REGULATOR**

Input Supply Voltage .....	+18V	Output Sinking Current (Terminal 7) .....	-10mA
Any Input or Output Voltage (Note 1) (Terminals 1, 2, 3, 5, 6, 7) .....	(GND - 0.3V) to (V <sub>IN</sub> <sup>+</sup> + 0.3V)	Power Dissipation (Note 2) Minidip .....	200mW
Output Source Current (Terminal 2) .....	50mA	TO-99 Can .....	300mW
(Terminal 3) .....	25mA		

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**OPERATING CHARACTERISTICS** V<sub>IN</sub><sup>+</sup> = 9V, V<sub>OUT</sub> = 5V, T<sub>A</sub> = +25°C, test circuit unless otherwise specified.

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
Input Voltage	V <sub>IN</sub> <sup>+</sup>	T <sub>A</sub> = +25°C 20°C ≤ T <sub>A</sub> ≤ +70°C	1.5 1.6		16.0 16.0	V
Quiescent Current	I <sub>Q</sub>	R <sub>L</sub> = ∞ 1.4V ≤ V <sub>OUT</sub> ≤ 8.5V V <sub>IN</sub> <sup>+</sup> = 16V V <sub>IN</sub> <sup>+</sup> = 9V		4.0 3.5	12 10	μA
Reference Voltage	V <sub>SET</sub>		1.2	1.3	1.4	V
Temperature Coefficient	$\frac{\Delta V_{SET}}{\Delta T}$	8.5V < V <sub>IN</sub> <sup>+</sup> < 9V		±200		ppm
Line Regulation	$\frac{\Delta V_{SET}}{V_{SET} \Delta V_{IN}}$	2V < V <sub>IN</sub> <sup>+</sup> < 15V		0.03		%/V
V <sub>SET</sub> Input Current	I <sub>SET</sub>			±0.01	10	nA
Shutdown Input Current	I <sub>SHDN</sub>			±0.01	10	nA
Shutdown Input Voltage	V <sub>SHDN</sub>	V <sub>SHDN</sub> HI: Both V <sub>OUT</sub> Disabled V <sub>SHDN</sub> LO: Both V <sub>OUT</sub> Enabled	1.4		0.3	V
Sense Pin Input Current	I <sub>SENSE</sub>			0.01	10	nA
Sense Pin Input Threshold Voltage	V <sub>CL</sub>	V <sub>CL</sub> = V <sub>OUT2</sub> - V <sub>SENSE</sub> (Current-Limit Threshold)		0.7		V
Input-Output Saturation Resistance (Note 3)	R <sub>SAT</sub>	V <sub>IN</sub> <sup>+</sup> = 2V V <sub>IN</sub> <sup>+</sup> = 9V V <sub>IN</sub> <sup>+</sup> = 15V		200 70 50		Ω
Load Regulation	$\frac{\Delta V_{OUT}}{\Delta I_{OUT}}$	ΔI <sub>OUT1</sub> = 100μA @ V <sub>OUT1</sub> = 5V ΔI <sub>OUT2</sub> = 10mA @ V <sub>OUT2</sub> = 5V		2.0 1.0		Ω
Available Output Current (V <sub>OUT2</sub> )	I <sub>OUT2</sub>	V <sub>IN</sub> <sup>+</sup> = 3V V <sub>OUT</sub> = V <sub>SET</sub> V <sub>IN</sub> <sup>+</sup> = 9V V <sub>OUT</sub> = 5V V <sub>IN</sub> <sup>+</sup> = 15V V <sub>OUT</sub> = 5V	10 25 40			mA
Negative-Tempco Output (Note 4)	V <sub>TC</sub>	Open-Circuit Voltage		0.9		V
	I <sub>TC</sub>	Maximum Sink Current	0	8	2.0	mA
Temperature Coefficient	$\frac{\Delta V_{TC}}{\Delta T}$	Open Circuit		+2.5		mV/°C
Minimum Load Current	I <sub>L(min)</sub>	(Includes V <sub>SET</sub> Divider)			1.0	μA

**Note 1:** Connecting any terminal to voltages greater than (V<sub>IN</sub><sup>+</sup> + 0.3V) or less than (GND - 0.3V) may cause destructive device latch-up. It is recommended that no inputs from sources operating on external power supplies be applied prior to ICL7663 power-up.

**Note 2:** Derate linearly above 50°C at 5mW/°C for minidip and 7.5mW/°C for TO-99 can.

**Note 3:** This parameter refers to the saturation resistance of the MOS pass transistor. The minimum input-output voltage differential at low current (under 5mA), can be determined by multiplying the load current (including set resistor current, but not quiescent current) by this resistance.

**Note 4:** This output has a positive temperature coefficient. Using it in combination with the inverting input of the regulator at V<sub>SET</sub>, a negative coefficient results in the output voltage. See Figure 3 for details. Pin will not source current.

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# ICL7663/7664



## ABSOLUTE MAXIMUM RATINGS, ICL7664 NEGATIVE REGULATOR

Input Supply Voltage .....	-18V	Power Dissipation (Note 2)	
Any Input or Output Voltage (Note 1)	(GND + 0.3V) to	Minidip .....	200mW
(Terminals 1, 2, 3, 5, 6, 7) .....	( $V_{IN}^- - 0.3V$ )	TO-99 Can .....	300mW
Output Sink Current			
(Terminals 1, 7) .....	-25mA		

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## OPERATING CHARACTERISTICS $V_{IN}^- = -9V$ , $V_{OUT} = -5V$ , $T_A = +25^\circ C$ , test circuit unless otherwise specified.

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
Input Voltage	$V_{IN}^-$	$T_A = +25^\circ C$ $0^\circ C \leq T_A \leq +70^\circ C$	-1.5 -1.6		-16.0 -16.0	V
Quiescent Current	$I_Q$	$\left\{ \begin{array}{l} R_L = \infty \\ -1.4V \leq V_{OUT} \leq -8.5V \end{array} \right\}$ $V_{IN}^- = 16V$ $V_{IN}^- = 9V$		4.0 3.5	12 10	$\mu A$
Reference Voltage	$V_{SET}$		-1.2	-1.3	-1.4	V
Temperature Coefficient	$\frac{\Delta V_{SET}}{\Delta T}$	$-8.5V < V_{IN}^- < -9V$		$\pm 200$		ppm
Line Regulation	$\frac{\Delta V_{SET}}{V_{SET} \Delta V_{IN}}$	$-2V < V_{IN}^- < -15V$		0.03		%/V
$V_{SET}$ Input Current	$I_{SET}$			$\pm 0.01$	10	nA
Shutdown Input Current	$I_{SHDN}$			$\pm 0.01$	10	nA
Shutdown Input Voltage	$V_{SHDN}$	$V_{SHDN} HI$ : Both $V_{OUT}$ Enabled $V_{SHDN} LO$ : Both $V_{OUT}$ Disabled	-0.3		-1.4	V
Sense Pin Input Current	$I_{SENSE}$			0.01	10	nA
Sense Pin Input Threshold Voltage	$V_{CL}$	$V_{CL} = V_{OUT2} - V_{SENSE}$ (Current-Limit Threshold)		-0.35		V
Input-Output Saturation Resistance (Note 3)	$R_{SAT}$	$V_{IN}^- = 2V$ $V_{IN}^- = 9V$ $V_{IN}^- = 15V$		150 40 30		$\Omega$
Load Regulation	$\frac{\Delta V_{OUT}}{\Delta I_{OUT}}$	$\Delta I_{OUT} = 100\mu A$ @ $V_{OUT} = -5V$		2.0		$\Omega$
Output Current, $V_{OUT1}$ or $V_{OUT2}$	$I_{OUT}$	$V_{IN}^- = 3V$ $V_{OUT} = V_{SET}$ $V_{IN}^- = 9V$ $V_{OUT} = -5V$ $V_{IN}^- = 15V$ $V_{OUT} = -5V$		-2 -20 -40		mA
Minimum Load Current (Includes $V_{SET}$ Divider)	$I_{L(min)}$				1.0	$\mu A$

**Note 1:** Connecting any terminal to voltages greater than (GND + 0.3V) or less than ( $V_{IN}^- - 0.3V$ ) may cause destructive device latch-up. It is recommended that no inputs from sources operating on external power supplies be applied prior to ICL7664 power-up.

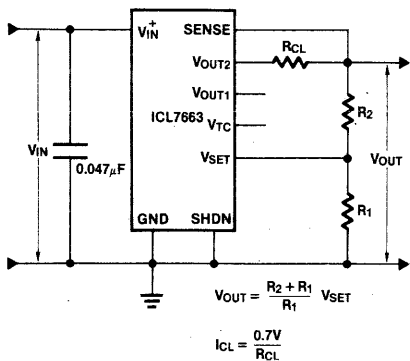
**Note 2:** Derate linearly above 50°C at 5mW/°C for minidip and 7.5mW/°C for TO-99 can.

**Note 3:** This parameter refers to the saturation resistance of the MOS pass transistor. The minimum input-output voltage differential can be determined by multiplying the load current (including set resistor current, but not quiescent current) by this resistance.

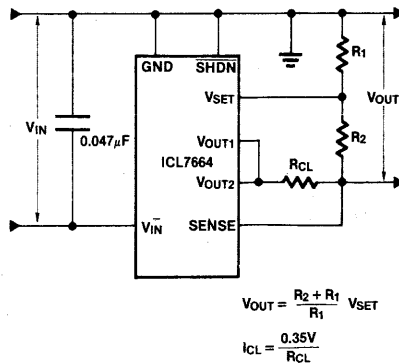
# ICL7663/7664



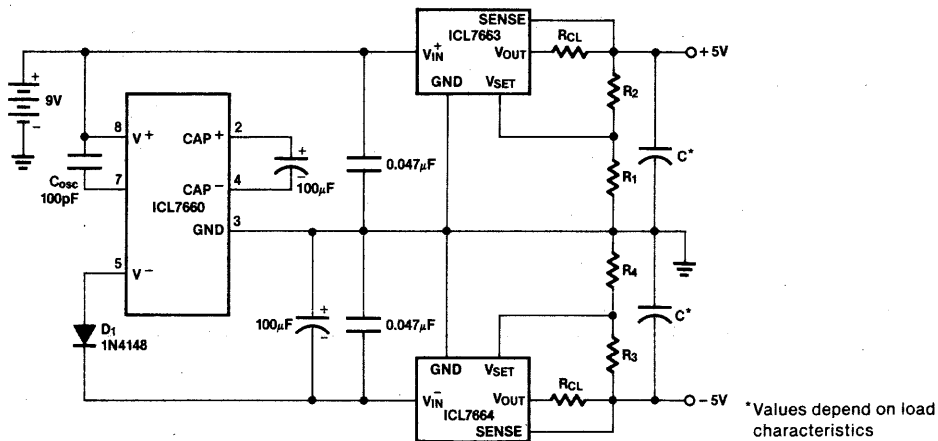
## APPLICATIONS



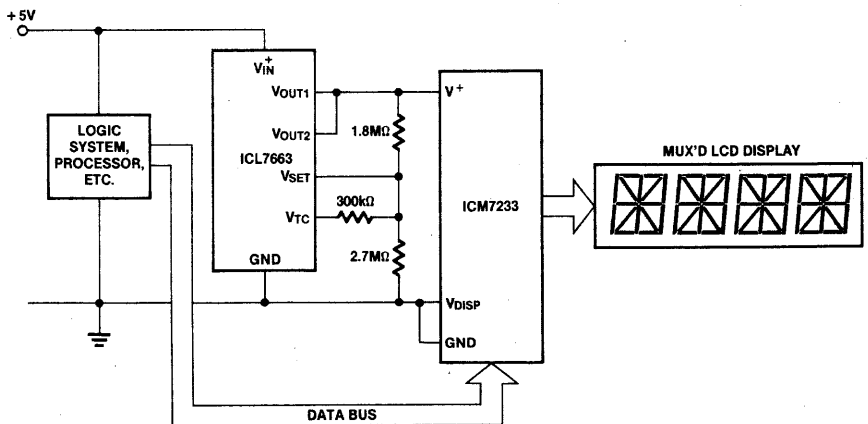
**Figure 4. Basic Application of ICL7663 as Positive Regulator with Current Limit**



**Figure 5. Basic Application of ICL7664 as Negative Regulator with Current Limit**



**Figure 6. Generating regulated split supplies from a single supply.** The oscillation frequency of the ICL7660 is reduced by the external oscillator capacitor, so that it inverts the battery voltage more efficiently.



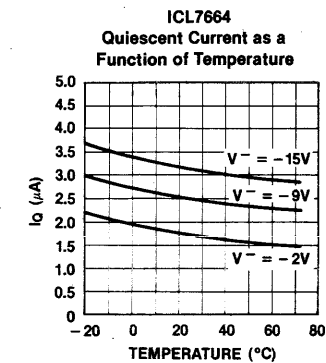
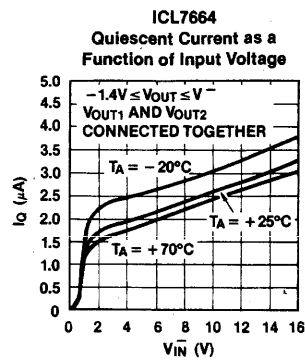
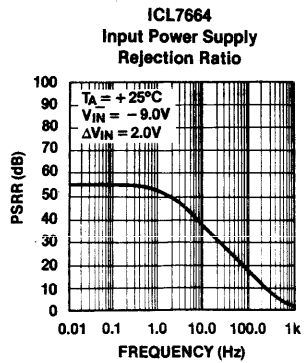
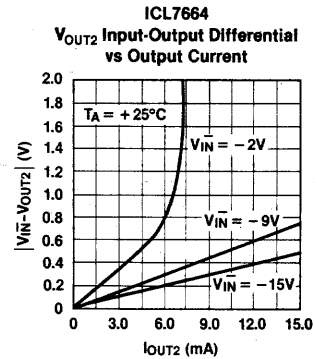
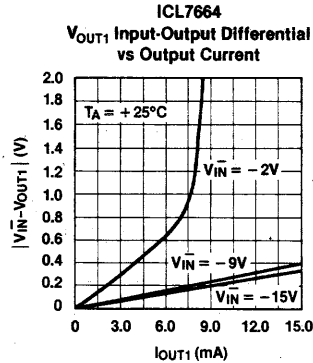
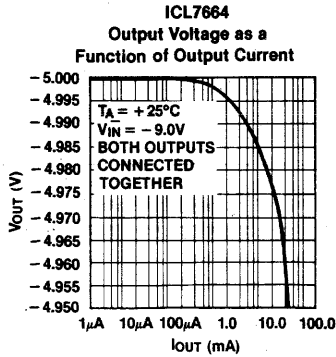
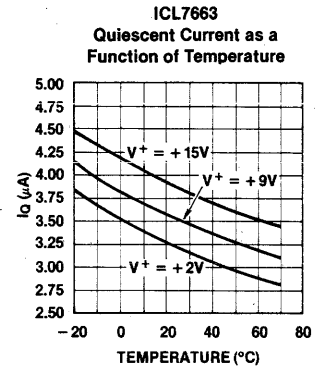
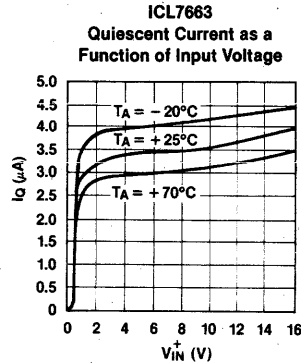
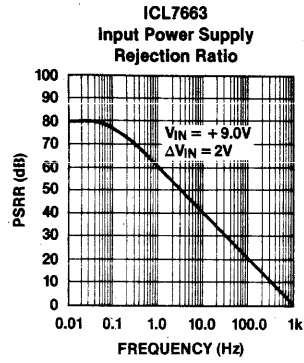
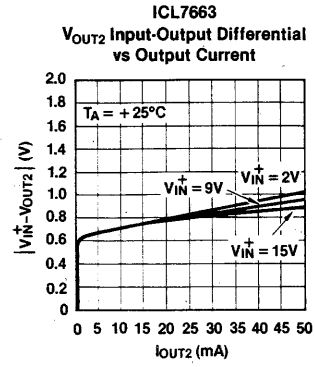
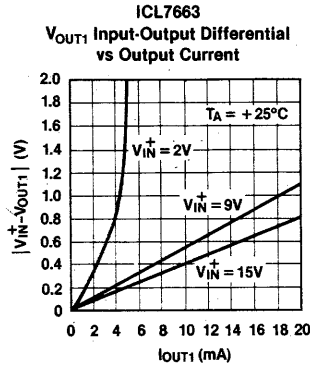
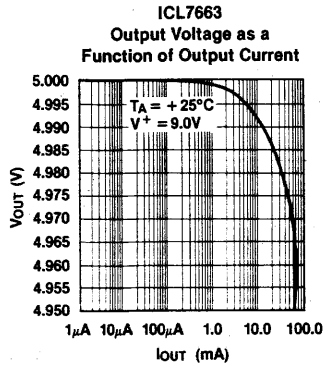
**Figure 7. Driving a Multiplexed LCD Display.** The negative temperature coefficient drive voltage to the displays allows consistent operation over more than 40°C temperature span, as opposed to about 10°C with a fixed drive voltage. Values based on EPSON LDB-728 display or similar.

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# ICL7663/7664

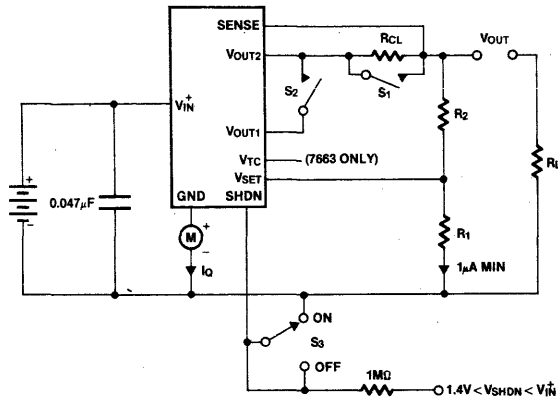
## TYPICAL CHARACTERISTICS



# ICL7663/7664



## TEST CIRCUIT



**Note 1:** S<sub>1</sub> when closed, disables output current limiting  
**Note 2:** For ICL7664, exchange V<sub>OUT1</sub> and V<sub>OUT2</sub>. S<sub>2</sub> action differs, as follows:

Device	S <sub>2</sub> Closed	S <sub>2</sub> Open
ICL7663	V <sub>OUT1</sub>	V <sub>OUT2</sub>
ICL7664	V <sub>OUT1</sub> + V <sub>OUT2</sub>	V <sub>OUT1</sub>

**Note 3:**  $V_{OUT} = \frac{R_2 + R_1}{R_1} V_{SET}$

**Note 4:** I<sub>Q</sub> quiescent current is measured at GND pin by meter M

**Note 5:** S<sub>3</sub> when ON, permits normal operation, when OFF, shuts down both V<sub>OUT1</sub> and V<sub>OUT2</sub>

Test Circuit for ICL7663/64 (Polarities shown are for ICL7663. Reverse for ICL7664)

## DETAILED DESCRIPTION

The ICL7663 and ICL7664 are CMOS integrated circuits which contain all the functions of a voltage regulator plus protection circuitry on a single monolithic chip. Referring to the block diagrams (Figures 1 and 2), each contains a bandgap-type voltage reference of 1.3 Volts; this voltage, therefore, is the lowest output voltage the regulators can control (–1.3V for the ICL7664). Error amplifier A drives either a P-channel (ICL7663) or an N-channel (ICL7664) pass transistor which is sufficient for low (under about 5mA) currents; this transistor is augmented by a duplicate in the ICL7664, which permits higher current outputs. In the ICL7663, the high current output is formed by an NPN transistor connected as a follower. This configuration gives more gain and lower output impedance.

Logic-controlled shutdown is implemented via an MOS transistor of the appropriate polarity. Current-sensing is achieved with comparator C, which functions with the V<sub>OUT2</sub> line on each chip. Finally, the positive regulator (ICL7663 only) has an output (V<sub>TC</sub>) from a buffer amplifier (B), which can be used to generate programmable-temperature-coefficient output voltages.

The amplifiers, reference and comparator circuitry all operate at bias levels well below 1µA to achieve the extremely low quiescent current. This does limit the dynamic response of the circuits, however, and transients are best dealt with outside the regulator loop.

## BASIC OPERATION

The ICL7663 and ICL7664 are designed to regulate battery voltages in the 5V to 15V region at maximum load currents of about 5mA to 30mA. Although intended as low power devices, power dissipation limits must be observed. For example, the power dissipation in the case of a 15V supply regulated down to 5V with a load current of 30mA clearly exceeds the power dissipation rating of the minidip: (15–5) (30) (10<sup>-3</sup>) = 300mW. The test circuit illustrates proper use of the devices. Although the following discussion refers to the ICL7663, it applies as well to the parallel features of the ICL7664 as long as the appropriate polarities are reversed. Individual features and precautions will be discussed where appropriate.

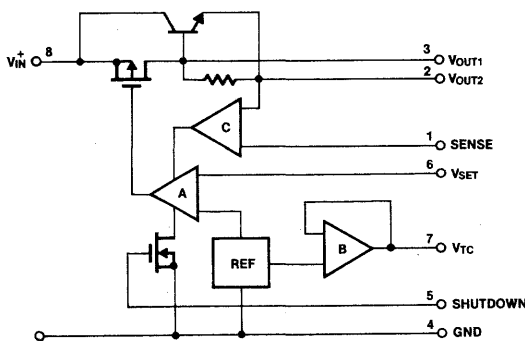


Figure 1. Block Diagram of the ICL7663

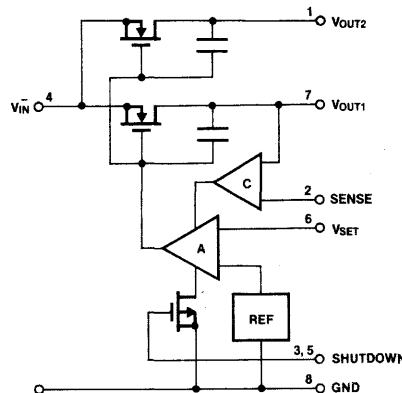


Figure 2. Block Diagram of the ICL7664

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## ICL7663/7664



CMOS devices generally require two precautions: every input pin must go somewhere, and maximum values of applied voltages and current limits must be rigorously observed. Neglecting these precautions may lead to, at the least, incorrect or non-operation, and at worst, destructive device failure. To avoid the problem of latchup, do not apply inputs to any pins before supply voltage is applied.

**Input Voltages**—These regulators accept working inputs of about 1.4V to 16V. When power is applied, the rate-of-rise of the input may be hundreds of volts per microsecond. This is potentially harmful to the regulators, where internal operating currents are in the nanoampere range. The 0.047 $\mu$ F capacitor on the device side of the switch will limit inputs to a safe level around 2V/ $\mu$ s. Use of this capacitor is suggested in all applications. In severe rate-of-rise cases, it may be advisable to use an RC network on the SHutDown pin to delay output turn-on. Battery charging surges, transients, and assorted noise signals should be kept from the regulators by RC filtering, zener protection, or even fusing.

**Output Voltages**—The resistor divider  $R_2/R_1$  is used to scale the reference voltage,  $V_{SET}$ , to the desired output using the formula  $V_{OUT} = (1 + R_2/R_1) V_{SET}$ . In the ICL7664,  $V_{IN}$  and  $V_{SET}$  are negative, so  $V_{OUT}$  will be also. Suitable arrangements of these resistors, using a potentiometer, enables exact values for  $V_{OUT}$  to be obtained. Because of the low leakage current of the  $V_{SET}$  terminal, these resistors can be tens of megohms for minimum additional quiescent drain current. However, some load current is required for proper operation, so for extremely low-drain applications it is necessary to draw at least 1 $\mu$ A. This can include the current for  $R_2$  and  $R_1$ .

Output voltages up to nearly the  $V_{IN}$  supply may be obtained at low load currents, while the low limit is the reference voltage. The minimum input-output differential in each regulator is obtained using the  $V_{OUT1}$  terminal.

**Output Currents**—For the ICL7663, low output currents of less than 5mA are obtained with the least input-output differential from the  $V_{OUT1}$  terminal (connect  $V_{OUT2}$  to  $V_{OUT1}$ ). Either output may be used on the ICL7664, with the unused output connected to  $V_{IN}$ . Where higher currents are needed, use  $V_{OUT2}$  on the ICL7663 ( $V_{OUT1}$  should be left open in this case) and parallel  $V_{OUT1}$  and  $V_{OUT2}$  on the ICL7664.

High output currents can be obtained only as far as package dissipation allows. It is strongly recommended that output current-limit sensing be used in such cases.

**Current-Limit Sensing**—The on-chip comparator (C in the block diagrams) permits shutdown of the regulator output in the event of excessive current drain. As the test circuits show, a current-limiting resistor,  $R_{CL}$ , is placed in series with  $V_{OUT2}$ , and the SENSE terminal is connected to the load side of  $R_{CL}$ . When the current through  $R_{CL}$  is high enough to produce a voltage drop equal to  $V_{CL}$  (0.7V for ICL7663, 0.35V for ICL7664) the voltage feedback is bypassed and the regulator output will be limited to this current. Therefore, when the maximum load current ( $I_{LOAD}$ ) is determined, simply divide  $V_{CL}$  by  $I_{LOAD}$  to obtain the value for  $R_{CL}$ .

**Logic-Controllable Shutdown**—When equipment is not needed continuously (e.g., in remote data-acquisition systems), it is desirable to eliminate its drain on the system until it is required. This usually means switches, with their unreliable contacts. Instead, the ICL7663 and ICL7664 can be shut down by a logic signal, leaving only  $I_Q$  (under 4 $\mu$ A) as a drain on the power source. Since this pin must not be left open, it should be tied to ground if not needed. A voltage of less than 0.3V for the ICL7663, and greater than -0.3V for the ICL7664 will keep the regulator ON, and a voltage level of more than 1.4V but less than  $V_{IN}^+$  for the ICL7663, and less than -1.4V but not less than  $V_{IN}^-$  for the ICL7664 control will turn the outputs OFF. If there is a possibility that the control signal could exceed the regulator input ( $V_{IN}^+$  or  $V_{IN}^-$ ), the current from this signal should be limited to 100 $\mu$ A maximum by a high-value (1M $\Omega$ ) series resistor. This situation may occur when the logic signal originates from a separately-powered system from that of the regulator.

**Additional Circuit Precautions**—These regulators have poor rejection of voltage fluctuations from AC sources above 10Hz or so. To prevent the output from responding (where this might be a problem), a reservoir capacitor across the load is advised. The value of this capacitor is chosen so that the regulated output voltage reaches 90% of its final value in 20ms. From

$$I = C \frac{\Delta V}{\Delta t}, C = I_{OUT} \frac{(20 \times 10^{-3})}{0.9 V_{OUT}} = 0.022 \frac{I_{OUT}}{V_{OUT}}$$

In addition, where such a capacitor is used, a current-limiting resistor is also suggested (see "Current-Limit Sensing").

**Producing Output Voltages With Negative Temperature Coefficients**—The ICL7663 has an additional output (not present on the ICL7664) which is 0.9V relative to GND and has a tempco of +2.5mV/ $^{\circ}$ C. By applying this voltage to the inverting input of amplifier A (i.e., the  $V_{SET}$  pin), output voltages having negative TC may be produced. The TC of the output voltage is controlled by the  $R_2/R_3$  ratio (see Figure 3 and its design equations).

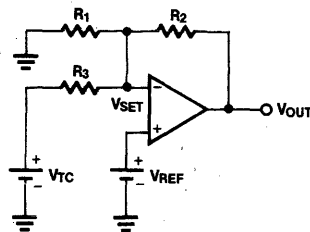


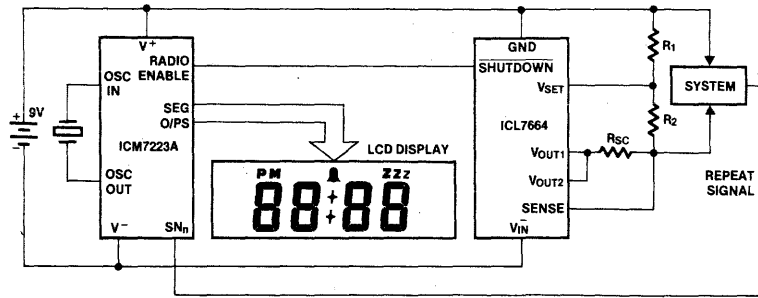
Figure 3. Generating Negative Temperature Coefficients

$$\text{EQ. 1: } V_{OUT} = V_{SET} \left( 1 + \frac{R_2}{R_1} \right) + \frac{R_2}{R_3} (V_{SET} - V_{TC})$$

$$\text{EQ. 2: } TC V_{OUT} = - \frac{R_2}{R_3} (TC V_{TC}) \text{ in mV}^{\circ}\text{C}$$

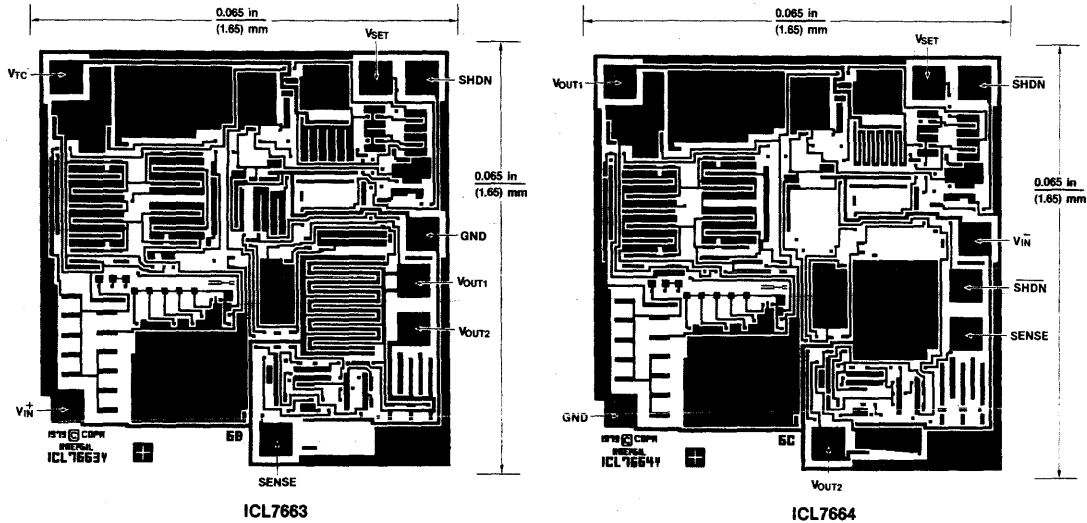
WHERE:  $V_{SET} = 1.3\text{V}$   
 $V_{TC} = 0.9\text{V}$   
 $TC V_{TC} = +2.5\text{mV}^{\circ}\text{C}$

APPLICATIONS (Continued)



**Figure 8. Once a Day System.** This circuit will turn on a regulated supply to a system for one minute every day, via the SHUTDOWN pin on the ICL7664, and under control of the ICM7223A Alarm Clock circuit. If the system decides it needs another one minute activation, pulling the REPEAT line to V<sup>+</sup> (GND) during one activation will trigger a subsequent activation after a snooze interval set by the choice of SN pins (2 mins shown). Alternatively, activation of the Sleep timer, without pause, can be achieved. See ICM7223A data sheet for details.

CHIP TOPOGRAPHIES



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**PRELIMINARY**  
 Applications Subject To Change Without Notice

# ICL7665

## Micropower Under-/Over-Voltage Detector

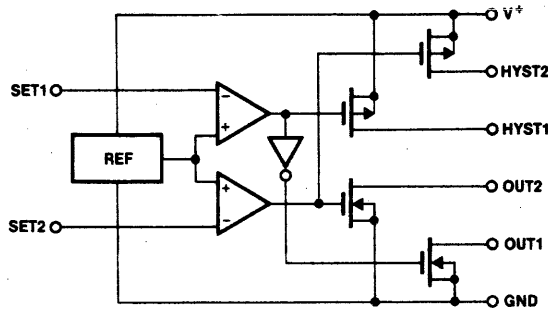
### FEATURES

- Exceptionally low supply current ( $\approx 3\mu\text{A}$  typ)
- Individually programmable upper and lower trip voltages and hysteresis levels
- Accurate on-chip bandgap reference, used by both detectors
- Up to 20mA output current sinking ability
- Wide supply voltage range

### GENERAL DESCRIPTION

The ICL7665 contains two individually programmable voltage detectors on a single chip. Requiring only  $\sim 3\mu\text{A}$  for operation, the device is intended for battery-operated systems and instruments which require high or low voltage warnings, settable trip points, or fault monitoring and correction. Typical applications are battery-backup computer memories, battery-operated medical devices, radiation dosimeters, pocket pagers, portable calibrators and test instruments, and charging systems.

### BLOCK DIAGRAM

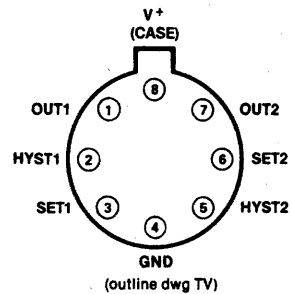
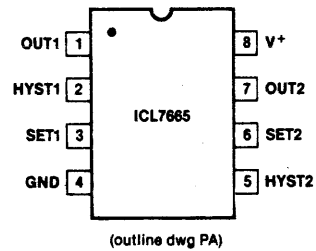


#### Conditions\*

- $V_{SET1} > 1.3\text{V}$ , OUT1 switch ON      HYST1 switch ON
- $V_{SET1} < 1.3\text{V}$ , OUT1 switch OFF    HYST1 switch OFF
- $V_{SET2} > 1.3\text{V}$ , OUT2 switch OFF    HYST2 switch ON
- $V_{SET2} < 1.3\text{V}$ , OUT2 switch ON    HYST2 switch OFF

\*See Operating Characteristics for exact thresholds.

### PIN CONFIGURATIONS



### ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
ICL7665PA	-20°C to +70°C	8 Lead MiniDIP
ICL7665TV	-20°C to +70°C	8 Lead TO-99
ICL7665/D	—	DICE Only

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**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage..... -0.3V to +18V  
 Output Voltages OUT1 and OUT2  
 (with respect to GND) (Note 2)..... -0.3V to +18V  
 Output Voltages HYST1 and HYST2  
 (with respect to V<sup>+</sup>) (Note 2)..... +0.3V to -18V  
 Input Voltages SET1 and SET2  
 (Note 2)..... (GND - 0.3V) to (V<sup>+</sup> + 0.3V)

Maximum Sink Output Current OUT1 and OUT2..... 25mA  
 Maximum Source Output Current  
 HYST1 and HYST2..... -25mA  
 Power Dissipation (Note 1)..... 200mW  
 Operating Temperature Range..... -20°C to +70°C  
 Storage Temperature Range..... -55°C to +125°C

**Note 1:** Derate above +25°C ambient temperature at 4mW/°C.

**Note 2:** Due to the SCR structure inherent in the CMOS process used to fabricate these devices, connecting any terminal to voltages greater than (V<sup>+</sup> + 0.3V) or less than (GND - 0.3V) may cause destructive device latchup. For this reason, it is recommended that no inputs from external sources not operating from the same power supply be applied to the device before its supply is established, and that in multiple supply systems, the supply to the ICL7665 be turned on first. If this is not possible, currents into inputs and/or outputs must be limited to ±0.5mA and voltages must not exceed those defined above.

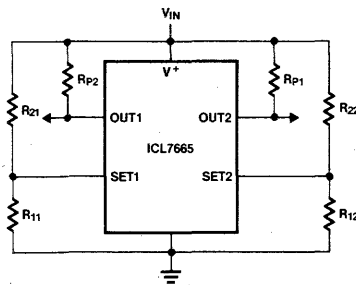
Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**DC OPERATING CHARACTERISTICS** (V<sup>+</sup> = 5V, T<sub>A</sub> = +25°C, test circuit unless otherwise specified.)

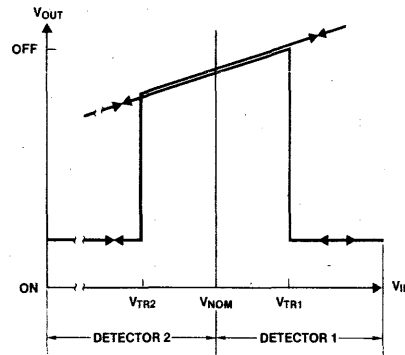
PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNITS
			MIN	TYP	MAX	
Operating Supply Voltage	V <sup>+</sup>	T <sub>A</sub> = +25°C -20°C ≤ T <sub>A</sub> ≤ +70°C	1.6 1.8		16.0 16.0	V
Supply Current	I <sup>+</sup>	GND ≤ V <sub>SET1</sub> , V <sub>SET2</sub> ≤ V <sup>+</sup> All Outputs Open Circuit V <sup>+</sup> = 2V V <sup>+</sup> = 9V V <sup>+</sup> = 15V		2.5 2.6 2.9	10 10 15	μA
Input Trip Voltage	V <sub>SET1</sub> V <sub>SET2</sub>		1.15 1.2	1.3 1.3	1.45 1.4	V
Temperature Coefficient of V <sub>SET</sub>	$\frac{\Delta V_{SET}}{\Delta T}$			200		ppm/°C
Supply Voltage Sensitivity of V <sub>SET1</sub> , V <sub>SET2</sub>	$\frac{\Delta V_{SET}}{\Delta V_S}$	R <sub>OUT1</sub> , R <sub>OUT2</sub> , R <sub>HYST1</sub> , R <sub>HYST2</sub> = 1 MΩ		0.004		%/V
Output Leakage Currents on OUT and HYST	I <sub>OLK</sub> I <sub>HLK</sub>	V <sub>SET</sub> = 0V or V <sub>SET</sub> ≥ 2V		10 -10	200 -100	nA
	I <sub>OLK</sub> I <sub>HLK</sub>	V <sup>+</sup> = 15V, T <sub>A</sub> = 70°C V <sup>+</sup> = 15V, T <sub>A</sub> = 70°C			2000 -500	
Output Saturation Voltages	V <sub>OUT1</sub> V <sub>OUT1</sub> V <sub>OUT1</sub> V <sub>HYST1</sub> V <sub>HYST1</sub> V <sub>HYST1</sub> V <sub>OUT2</sub> V <sub>OUT2</sub> V <sub>OUT2</sub> V <sub>HYST2</sub> V <sub>HYST2</sub> V <sub>HYST2</sub>	V <sup>+</sup> = 2V, V <sub>SET1</sub> = 2V, I <sub>OUT1</sub> = 2mA V <sup>+</sup> = 5V, V <sub>SET1</sub> = 2V, I <sub>OUT1</sub> = 2mA V <sup>+</sup> = 15V, V <sub>SET1</sub> = 2V, I <sub>OUT1</sub> = 2mA V <sup>+</sup> = 2V, V <sub>SET1</sub> = 2V, I <sub>HYST1</sub> = -0.5mA V <sup>+</sup> = 5V, V <sub>SET1</sub> = 2V, I <sub>HYST1</sub> = -0.5mA V <sup>+</sup> = 15V, V <sub>SET1</sub> = 2V, I <sub>HYST1</sub> = -0.5mA V <sup>+</sup> = 2V, V <sub>SET2</sub> = 0V, I <sub>OUT2</sub> = 2mA V <sup>+</sup> = 5V, V <sub>SET2</sub> = 0V, I <sub>OUT2</sub> = 2mA V <sup>+</sup> = 15V, V <sub>SET2</sub> = 0V, I <sub>OUT2</sub> = 2mA V <sup>+</sup> = 2V, V <sub>SET2</sub> = 2V, I <sub>HYST2</sub> = -0.2mA V <sup>+</sup> = 5V, V <sub>SET2</sub> = 2V, I <sub>HYST2</sub> = -0.5mA V <sup>+</sup> = 15V, V <sub>SET2</sub> = 2V, I <sub>HYST2</sub> = -0.5mA		0.2 0.1 0.06 -0.15 -0.05 -0.02 0.2 0.15 0.11 -0.25 -0.43 -0.35	0.5 0.3 0.2 -0.3 -0.15 -0.10 0.5 0.3 0.25 -0.8 -1.0 -0.8	V
V <sub>SET</sub> Input Leakage Current	I <sub>SET</sub>	GND ≤ V <sub>SET</sub> ≤ V <sup>+</sup>		0.01	10	nA
ΔV <sub>SET</sub> Input for Complete Output Change	ΔV <sub>SET</sub>	R <sub>OUT</sub> = 4.7kΩ, R <sub>HYST</sub> = 20kΩ V <sub>OUTLO</sub> = 1% V <sup>+</sup> , V <sub>OUTH</sub> = 99% V <sup>+</sup>		1		mV
Difference in Trip Voltages	V <sub>SET1</sub> - V <sub>SET2</sub>	R <sub>OUT</sub> , R <sub>HYST</sub> = 1MΩ		±5	±50	
Output/Hysteresis Difference		R <sub>OUT</sub> , R <sub>HYST</sub> = 1MΩ		±1		

LINEAR Intersil

**ICL7665**  
**APPLICATIONS**



(a) Circuit Configuration



(b) Transfer Characteristics

Figure 1. Simple Threshold Detector

Figure 1 shows the simplest connection of the ICL7665 for threshold detection. From the graph (b), it can be seen that at low input voltages OUT1 is OFF, or high, while OUT2 is ON, or low. As the input rises (e.g. at power-on) toward  $V_{NOM}$  (usually the eventual operating voltage), OUT2 goes high on reaching  $V_{TR2}$ . If the voltage rises above  $V_{NOM}$  as much as  $V_{TR1}$ , OUT1 goes low. The equations giving  $V_{SET1}$  and  $V_{SET2}$  are, from Figure 1(a):

$$V_{SET1} = V_{IN} \frac{R_{11}}{(R_{11} + R_{21})} \quad V_{SET2} = V_{IN} \frac{R_{12}}{(R_{12} + R_{22})}$$

Since the voltage to trip each comparator is nominally 1.3V, the value of  $V_{IN}$  for each trip point can be found from

$$V_{TR1} = V_{SET1} \frac{(R_{11} + R_{21})}{R_{11}} = 1.3 \frac{(R_{11} + R_{21})}{R_{11}} \text{ for detector 1 and}$$

$$V_{TR2} = V_{SET2} \frac{(R_{12} + R_{22})}{R_{12}} = 1.3 \frac{(R_{12} + R_{22})}{R_{12}} \text{ for detector 2.}$$

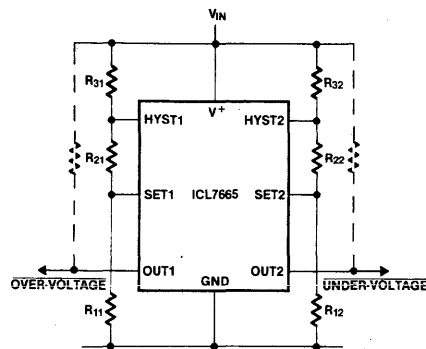
Either detector may be used alone, as well as both together, in any of the circuits shown here.

When  $V_{IN}$  is very close to one of the trip voltages, normal variations and noise may cause it to wander back and forth across this level, leading to erratic output ON and OFF condi-

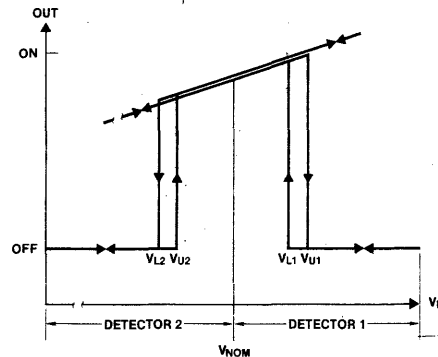
tions. The addition of hysteresis, making the trip points slightly different for rising and falling inputs, will avoid this condition.

Figure 2(a) shows how to set up such hysteresis, while Figure 2(b) shows how the hysteresis around each trip point produces switching action at different points depending on whether  $V_{IN}$  is rising or falling (the arrows indicate direction of change). The HYST outputs are basically switches which short out  $R_{31}$  or  $R_{32}$  when  $V_{IN}$  is above the respective trip point. Thus if the input voltage rises from a low value, the trip point will be controlled by  $R_{1n}$ ,  $R_{2n}$  and  $R_{3n}$ , until the trip point is reached. As this value is passed, the detector changes state,  $R_{3n}$  is shorted out, and the trip point becomes controlled by only  $R_{1n}$  and  $R_{2n}$ , a lower value. The input will then have to fall to this new point to restore the initial comparator state, but as soon as this occurs, the trip point will be raised again.

An alternative circuit for obtaining hysteresis is shown in Figure 3. In this configuration, the HYST pins put the extra resistor in parallel with the upper setting resistor. The values of the resistors differ, but the action is essentially the same. The governing equations are given in Table 1. These ignore the effects of the resistance of the HYST outputs, but these can normally be neglected if the resistor values are above about 100k $\Omega$ .



(a) Circuit Configuration



(b) Transfer Characteristics

Figure 2. Threshold Detector with Hysteresis



# ICL8021—ICL8023 Low Power Operational Amplifiers

## FEATURES

- $\Delta V_{os} = 3 \text{ mV max}$  (adjustable to zero).
- $\pm 1\text{V}$  to  $\pm 18\text{V}$  Power Supply Operation.
- Power Consumption —  $20 \mu\text{W}$  @  $\pm 1\text{V}$ .
- Input Bias Current —  $30 \text{ nA max}$ .
- Internal Compensation.
- Pin-For-Pin Compatible With 741.
- Short Circuit Protected.

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage	$\pm 18\text{V}$
Differential Input Voltage (Note 1)	$\pm 15\text{V}$
Common Mode Input Voltage (Note 1)	$\pm 15\text{V}$
Output Short Circuit Duration	Indefinite
Power Dissipation (Note 2)	$300 \text{ mW}$
Operating Temperature Range	
8021M	$-55^\circ\text{C}$ to $+125^\circ\text{C}$
8021C	$0^\circ\text{C}$ to $+70^\circ\text{C}$
Storage Temperature Range	$-65^\circ\text{C}$ to $+150^\circ\text{C}$
Lead Temperature (Soldering, 10 sec)	$+300^\circ\text{C}$

**NOTE 1:** For supply voltages less than  $\pm 15\text{V}$ , the absolute maximum input voltage is equal to the supply voltage.

**NOTE 2:** Rating applies for case temperatures to  $+125^\circ\text{C}$ ; derate linearly at  $5.6 \text{ mW}/^\circ\text{C}$  for ambient temperatures above  $+95^\circ\text{C}$ .

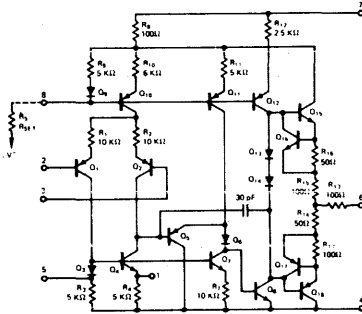
## GENERAL DESCRIPTION

The Intersil 8021 integrated circuit is a low power operational amplifier specifically designed for applications requiring very low standby power consumption over a wide range of supply voltages. The electrical characteristics of the 8021 can be tailored to a particular application by adjusting an external resistor,  $R_{SET}$ , which controls the quiescent current. This is advantageous because  $I_Q$  can be made independent of the supply voltages: it can be set to an extremely low value where power is critical, or to a larger value for high slew rate or wideband applications.

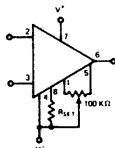
Other features of the 8021 include low input current that remains constant with temperature, low noise, high input impedance, internal compensation and pin-for-pin compatibility with the 741.

The Intersil 8022 (8023) consists of two (three) low power operational amplifiers in a single 14-pin DIP. Each amplifier is identical to an 8021 low power op amp, and has separate connections for adjusting its electrical characteristics by means of an external resistor,  $R_{SET}$ , which controls the quiescent current of that amplifier.

## SCHEMATIC DIAGRAM



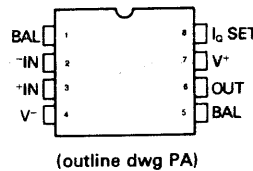
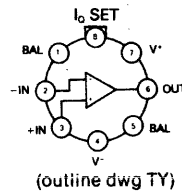
## VOLTAGE OFFSET NULL CIRCUIT



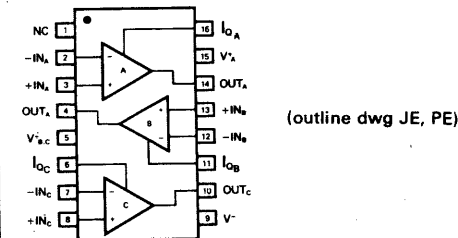
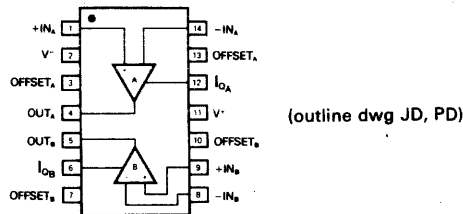
## ORDERING INFORMATION

ICL8021	C	TY	Package	
Basic			TY — TO-99 Metal Can	8021 only
Part Number			PA — 8 pin Minidip	
8021—Single			JD — 14 pin CERDIP	8022 only
8022—Dual			PD — 14 pin Plastic DIP	
8023—Triple			JE — 16 pin CERDIP	8023 only
			PE — 16 pin Plastic DIP	
			Temperature	
			C — Commercial — $0^\circ\text{C}$ to $70^\circ\text{C}$	
			M — Military — $-55^\circ\text{C}$ to $+125^\circ\text{C}$	

## PIN CONFIGURATIONS



**NOTE:** Pin 4 connected to case.



ELECTRICAL CHARACTERISTICS ( $V_S = \pm 6V$ , $I_Q = 30 \mu A$ , unless otherwise specified.)																																																	
CHARACTERISTICS	CONDITIONS	8021M			8021C			UNITS																																									
		MIN	TYP	MAX	MIN	TYP	MAX																																										
The following specifications apply for $T_A = 25^\circ C$ :																																																	
Input Offset Voltage	$R_S \leq 100 k\Omega$	2	3		2	6	mV																																										
Input Offset Current		.5	7.5		.7	10	nA																																										
Input Bias Current		5	20		7	30	nA																																										
Input Resistance		3	10		3	10	$M\Omega$																																										
Input Voltage Range	$V_S = \pm 15V$	$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$	V																																										
Common Mode Rejection Ratio	$R_S \leq 10 k\Omega$	70	80		70	80	dB																																										
Supply Voltage Rejection Ratio	$R_S \leq 10 k\Omega$	30	150		30	150	$\mu V/V$																																										
Output Resistance	Open Loop		2			2	$k\Omega$																																										
Output Voltage Swing	$R_L \geq 20 k\Omega$ , $V_S = \pm 15V$	$\pm 12$	$\pm 14$		$\pm 12$	$\pm 14$	V																																										
	$R_L \geq 10 k\Omega$ , $V_S = \pm 15V$	$\pm 11$	$\pm 13$		$\pm 11$	$\pm 13$	V																																										
Output Short-Circuit Current			$\pm 13$			$\pm 13$	mA																																										
Power Consumption	$V_{OUT} = 0$	360	480		360	600	$\mu W$																																										
Slew Rate (Unity Gain)			0.16			0.16	V/ $\mu s$																																										
Unity Gain Bandwidth	$R_L = 20 k\Omega$ , $V_{IN} = 20 mV$		270			270	kHz																																										
Transient Response (Unity Gain)	$R_L = 20 k\Omega$ , $V_{IN} = 20 mV$		1.3			1.3	$\mu s$																																										
		Risetime																																															
		Overshoot		10			10	%																																									
The following specifications apply for $0^\circ C \leq T_A < +70^\circ C$ (8021C) $-55^\circ C < +125^\circ C$ (8021M)																																																	
Input Offset Voltage	$R_S \leq 10 k\Omega$	2.0	4.0		2.0	7.5	mV																																										
Input Offset Current		1.0	11		1.5	15	nA																																										
Input Bias Current		10	32		15	50	nA																																										
Average Temperature Coefficient of Input Offset Voltage	$R_S \leq 10 k\Omega$	5			5		$\mu V/^\circ C$																																										
Average Temperature Coefficient of Input Offset Current		1.7			0.8		$pA/^\circ C$																																										
Large Signal Voltage Gain	$R_L = 10 k\Omega$	50	200		50	200	V/mV																																										
Output Voltage Swing	$R_L \geq 10 k\Omega$	$\pm 10$	$\pm 13$		$\pm 10$	$\pm 13$	V																																										
<b>QUIESCENT CURRENT ADJUSTMENT</b>																																																	
<p>QUIESCENT CURRENT SETTING RESISTOR (PIN 8 TO <math>V^-</math>)</p> <table border="1"> <thead> <tr> <th><math>V_S</math></th> <th><math>I_Q</math></th> <th>10 <math>\mu A</math></th> <th>30 <math>\mu A</math></th> <th>100 <math>\mu A</math></th> <th>300 <math>\mu A</math></th> </tr> </thead> <tbody> <tr> <td><math>\pm 15</math></td> <td></td> <td>1.5 <math>M\Omega</math></td> <td>470 <math>k\Omega</math></td> <td>150 <math>k\Omega</math></td> <td></td> </tr> <tr> <td><math>\pm 12</math></td> <td></td> <td>3.3 <math>M\Omega</math></td> <td>1.1 <math>M\Omega</math></td> <td>330 <math>k\Omega</math></td> <td>100 <math>k\Omega</math></td> </tr> <tr> <td><math>\pm 9</math></td> <td></td> <td>7.5 <math>M\Omega</math></td> <td>2.7 <math>M\Omega</math></td> <td>750 <math>k\Omega</math></td> <td>220 <math>k\Omega</math></td> </tr> <tr> <td><math>\pm 6</math></td> <td></td> <td>13 <math>M\Omega</math></td> <td>4 <math>M\Omega</math></td> <td>1.3 <math>M\Omega</math></td> <td>350 <math>k\Omega</math></td> </tr> <tr> <td><math>\pm 3</math></td> <td></td> <td>18 <math>M\Omega</math></td> <td>5.6 <math>M\Omega</math></td> <td>1.5 <math>M\Omega</math></td> <td>510 <math>k\Omega</math></td> </tr> <tr> <td><math>\pm 1.5</math></td> <td></td> <td>22 <math>M\Omega</math></td> <td>7.5 <math>M\Omega</math></td> <td>2.2 <math>M\Omega</math></td> <td>620 <math>k\Omega</math></td> </tr> </tbody> </table>								$V_S$	$I_Q$	10 $\mu A$	30 $\mu A$	100 $\mu A$	300 $\mu A$	$\pm 15$		1.5 $M\Omega$	470 $k\Omega$	150 $k\Omega$		$\pm 12$		3.3 $M\Omega$	1.1 $M\Omega$	330 $k\Omega$	100 $k\Omega$	$\pm 9$		7.5 $M\Omega$	2.7 $M\Omega$	750 $k\Omega$	220 $k\Omega$	$\pm 6$		13 $M\Omega$	4 $M\Omega$	1.3 $M\Omega$	350 $k\Omega$	$\pm 3$		18 $M\Omega$	5.6 $M\Omega$	1.5 $M\Omega$	510 $k\Omega$	$\pm 1.5$		22 $M\Omega$	7.5 $M\Omega$	2.2 $M\Omega$	620 $k\Omega$
$V_S$	$I_Q$	10 $\mu A$	30 $\mu A$	100 $\mu A$	300 $\mu A$																																												
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$\pm 6$		13 $M\Omega$	4 $M\Omega$	1.3 $M\Omega$	350 $k\Omega$																																												
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<p>QUIESCENT CURRENT SETTING RESISTOR (PIN 8 TO <math>V^-</math>)</p>																																																	

Intersil LINEAR



# ICL8038 Precision Waveform Generator/Voltage Controlled Oscillator

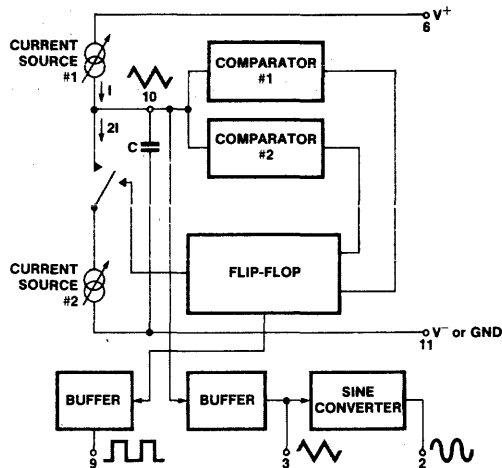
## FEATURES

- Low frequency drift with temperature - 50ppm/°C
- Simultaneous sine, square, and triangle wave outputs
- Low distortion - 1% (sine wave output)
- High linearity - 0.1% (triangle wave output)
- Wide operating frequency range - 0.001Hz to 0.3MHz
- Variable duty cycle - 2% to 98%
- High level outputs - TTL to 28V
- Easy to use - just a handful of external components required

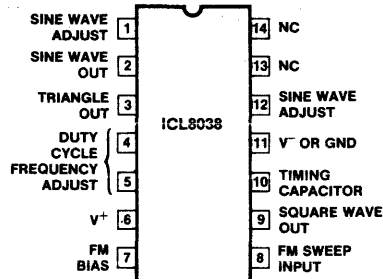
## GENERAL DESCRIPTION

The ICL8038 Waveform Generator is a monolithic integrated circuit capable of producing high accuracy sine, square, triangular, sawtooth and pulse waveforms with a minimum of external components. The frequency (or repetition rate) can be selected externally from .001Hz to more than 300kHz using either resistors or capacitors, and frequency modulation and sweeping can be accomplished with an external voltage. The ICL8038 is fabricated with advanced monolithic technology, using Schottky-barrier diodes and thin film resistors, and the output is stable over a wide range of temperature and supply variations. These devices may be interfaced with phase locked loop circuitry to reduce temperature drift to less than 50ppm/°C.

## BLOCK DIAGRAM



## PIN CONFIGURATION (outline dwg JD)



## ORDERING INFORMATION

TYPE	TEMPERATURE RANGE	STABILITY	PACKAGE	ORDER PART NUMBER
8038 CC	0°C to +70°C	250ppm/°C typ	CERDIP	ICL8038 CC JD
8038 BC	0°C to +70°C	150ppm/°C max	CERDIP	ICL8038 BC JD
8038 AC	0°C to +70°C	80ppm/°C max	CERDIP	ICL8038 AC JD
8038 BM	-55°C to +125°C	150ppm/°C max	CERDIP	ICL8038 BM JD
8038 AM	-55°C to +125°C	80ppm/°C max	CERDIP	ICL8038 AM JD



# ICL8038



## MAXIMUM RATINGS

Supply Voltage	±18V or 36V Total
Power Dissipation <sup>(1)</sup>	750mW
Input Voltage (any pin)	Not To Exceed Supply Voltages
Input Current (Pins 4 and 5)	25mA
Output Sink Current (Pins 3 and 9)	25mA
Storage Temperature Range	-65°C to +125°C
Operating Temperature Range:	
8038AM, 8038BM	-55°C to +125°C
8038AC, 8038BC, 8038CC	0°C to +70°C
Lead Temperature (Soldering, 10 sec.)	300°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**NOTE 1:** Derate ceramic package at 12.5mW/°C for ambient temperatures above 100°C.

## ELECTRICAL CHARACTERISTICS

(V<sub>SUPP</sub> = ±10V or +20V, T<sub>A</sub> = 25°C, R<sub>L</sub> = 10kΩ, Test Circuit Unless Otherwise Specified)

SYMBOL	GENERAL CHARACTERISTICS	8038CC			8038BC(BM)			8038AC(AM)			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>SUPP</sub>	Supply Voltage Operating Range										
V <sup>+</sup>	Single Supply	+10		+30	+10		30	+10		30	V
V <sup>+</sup> , V <sup>-</sup>	Dual Supplies	±5		±15	±5		±15	±5		±15	V
I <sub>SUPP</sub>	Supply Current (V <sub>SUPP</sub> = ±10V) <sup>(2)</sup>										
	8038AM, 8038BM					12	15		12	15	mA
	8038AC, 8038BC, 8038CC		12	20		12	20		12	20	mA
<b>FREQUENCY CHARACTERISTICS</b> (all waveforms)											
f <sub>max</sub>	Maximum Frequency of Oscillation	100,000			100,000			100,000			Hz
f <sub>sweep</sub>	Sweep Frequency of FM		10			10			10		kHz
	Sweep FM Range <sup>(3)</sup>		35:1			35:1			35:1		
	FM Linearity 10:1 Ratio		0.5			0.2			0.2		%
Δf/ΔT	Frequency Drift With Temperature <sup>(5)</sup> +25°C to +70°C (+125°C) 0°C (-40°C) to +25°C		250			150			80		ppm/°C
			250			200			120		
Δf/ΔV	Frequency Drift With Supply Voltage (Over Supply Voltage Range)		0.05			0.05			0.05		%/V <sub>SUPP</sub>
	Recommended Programming Resistors (R <sub>A</sub> and R <sub>B</sub> )	1000		1M	1000		1M	1000		1M	Ω
<b>OUTPUT CHARACTERISTICS</b>											
I <sub>OLK</sub>	<b>Square-Wave</b> Leakage Current (V <sub>g</sub> = 30V)			1			1			1	μA
V <sub>SAT</sub>	Saturation Voltage (I <sub>SINK</sub> = 2mA)		0.2	0.5		0.2	0.4		0.2	0.4	V
t <sub>r</sub>	Rise Time (R <sub>L</sub> = 4.7kΩ)		180			180			180		ns
t <sub>f</sub>	Fall Time (R <sub>L</sub> = 4.7kΩ)		40			40			40		ns
	Duty Cycle Adjust	2		98	2		98	2		98	%
	<b>Triangle/Sawtooth/Ramp</b> Amplitude (R <sub>TRI</sub> = 100kΩ)	0.30	0.33		0.30	0.33		0.30	0.33		xV <sub>SUPP</sub>
	Linearity		0.1			0.05			0.05		%
Z <sub>OUT</sub>	Output Impedance (I <sub>OUT</sub> = 5mA)		200			200			200		Ω
	<b>Sine-Wave</b> Amplitude (R <sub>SINE</sub> = 100kΩ)	0.2	0.22		0.2	0.22		0.2	0.22		xV <sub>SUPP</sub>
	THD (R <sub>S</sub> = 1MΩ) <sup>(4)</sup>		2.0	5		1.5	3		1.0	1.5	%
	THD Adjusted (Use Fig. 8b)		1.5			1.0			0.8		%

**NOTE 2:** R<sub>A</sub> and R<sub>B</sub> currents not included.

**NOTE 3:** V<sub>SUPP</sub> = 20V; R<sub>A</sub> and R<sub>B</sub> = 10kΩ, f ≈ 9kHz; Can be extended to 1000.1. See Figures 13 and 14.

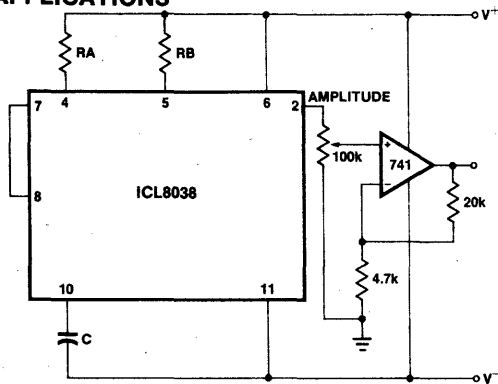
**NOTE 4:** 82kΩ connected between pins 11 and 12, Triangle Duty Cycle set at 50%. (Use R<sub>A</sub> and R<sub>B</sub>.)

**NOTE 5:** Fig. 2, pins 7 and 8 connected, V<sub>SUPP</sub> = ±10V. See Fig. 6c for T.C. vs V<sub>SUPP</sub>.

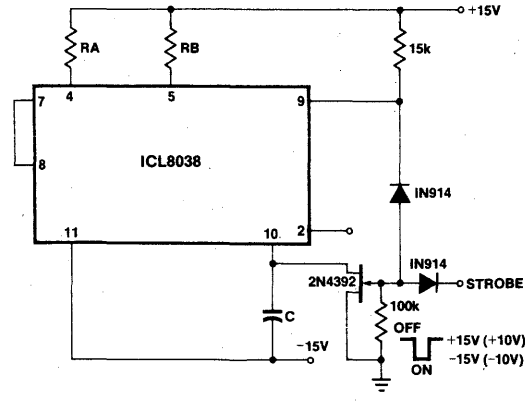
# ICL8038



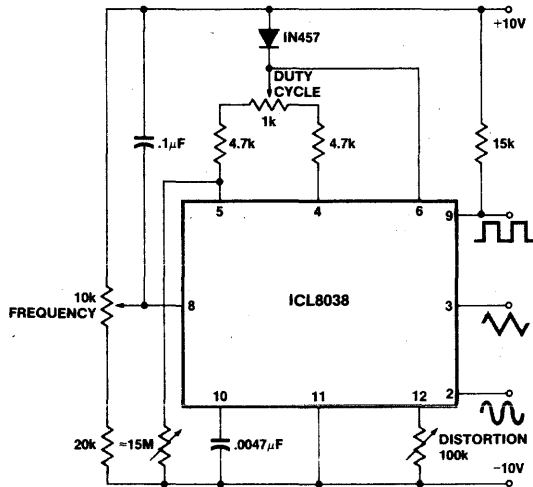
## APPLICATIONS



**Figure 4: Sine Wave Output Buffer Amplifiers.** The sine wave output has a relatively high output impedance (1kΩ Typ). The circuit of Figure 4 provides buffering, gain and amplitude adjustment. A simple op amp follower could also be used.

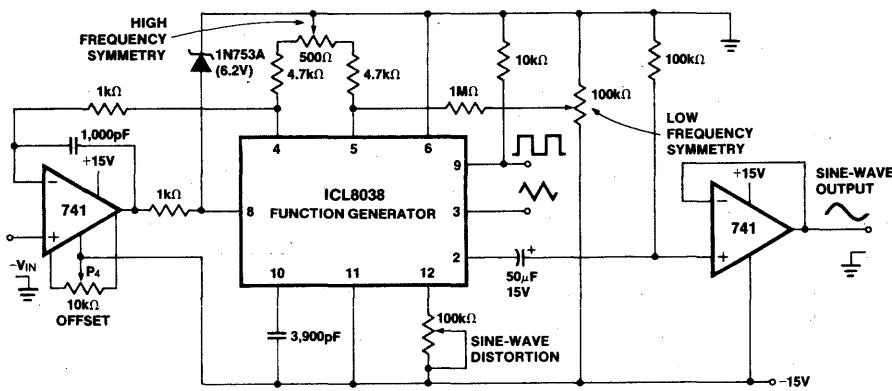


**Figure 5: Strobe-Tone Burst Generator.** With a dual supply voltage the external capacitor on Pin 10 can be shorted to ground to halt the 8038 oscillation. Figure 5 shows a FET switch, diode ANDED with an input strobe signal to allow the output to always start on the same slope.



**Figure 6: Variable Audio Oscillator, 20Hz to 20kHz.**

To obtain a 1000:1 Sweep Range on the 8038 the voltage across external resistors RA and RB must decrease to nearly zero. This requires that the highest voltage on control Pin 8 exceed the voltage at the top of RA and RB by a few hundred millivolts. The Circuit of Figure 6 achieves this by using a diode to lower the effective supply voltage on the 8038. The large resistor on pin 5 helps reduce duty cycle variations with sweep.



**Figure 7: Linear Voltage Controlled Oscillator**

The linearity of input sweep voltage versus output frequency can be significantly improved by using an op amp as shown in Figure 7.

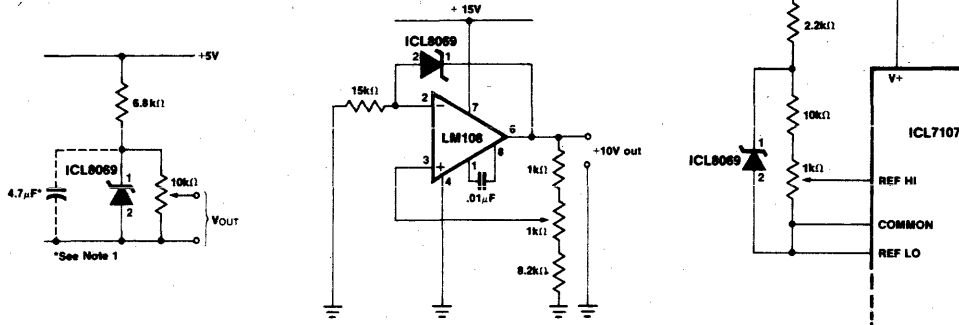
## FEATURES

- Temperature Coefficient guaranteed to 10 ppm/°C max.
- Low Bias Current . . . 50μA min
- Low Dynamic Impedance
- Low Reverse Voltage
- Low Cost

## GENERAL DESCRIPTION

The ICL8069 is a 1.2V temperature compensated voltage reference. It uses the band-gap principle to achieve excellent stability and low noise at reverse currents down to 50μA. Applications include analog-to-digital converters, digital-to-analog converters, threshold detectors, and voltage regulators. Its low power consumption makes it especially suitable for battery operated equipment.

## TYPICAL CONNECTION DIAGRAMS



(a) Simple Reference (1.2 volts or less)

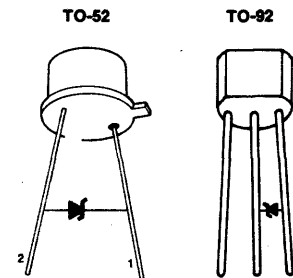
(b) Buffered 10V Reference using a single supply.

(c) Double regulated 100mV reference for ICL7107 one-chip DPM circuit.

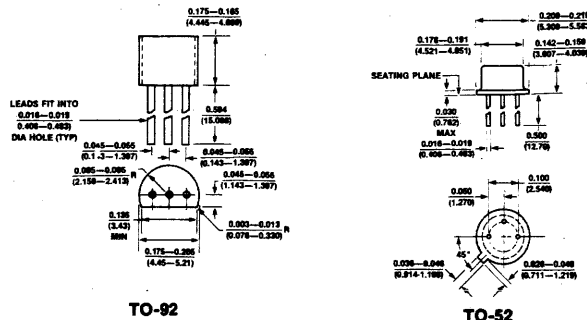
## ORDERING INFORMATION

Max. Temp. Coeff. of $V_{REF}$	Temp. Range	Order P/N TO-92	Order P/N TO-52
0.001%/°C	0°C to +70°C	—	ICL8069ACSQ
0.0025%/°C	0°C to +70°C	—	ICL8069BCSQ
0.005%/°C	0°C to +70°C	ICL8069CCZR	ICL8069CCSQ
0.005%/°C	-55°C to +125°C	—	ICL8069CMSQ
0.01%/°C	0°C to +70°C	ICL8069DCZR	ICL8069DCSQ
0.01%/°C	-55°C to +125°C	—	ICL8069DMSQ

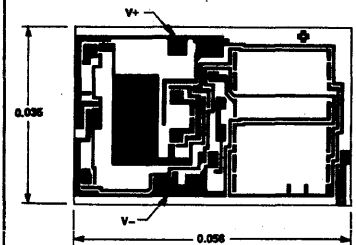
## PIN CONFIGURATION



## PACKAGE DIMENSIONS



## CHIP TOPOGRAPHY



### FEATURES

- High accuracy voltage sensing and generation: internal reference 1.15 volts typical
- Low sensitivity to supply voltage and temperature variations
- Wide supply voltage range: Typ. 1.8 to 30 volts
- Essentially constant supply current over full supply voltage range
- Easy to set hysteresis voltage range
- Defined output current limit - ICL8211  
High output current capability - ICL8212

### GENERAL DESCRIPTION

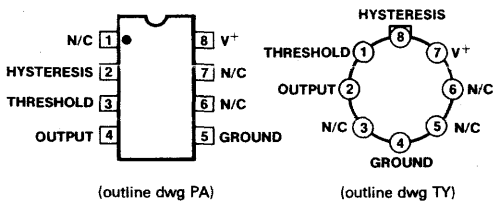
The Intersil ICL8211/12 are micropower bipolar monolithic integrated circuits intended primarily for precise voltage detection and generation. These circuits consist of an accurate voltage reference, a comparator and a pair of output buffer/drivers.

Specifically, the ICL8211 provides a 7mA current limited output sink when the voltage applied to the 'THRESHOLD' terminal is less than 1.15 volts (the internal reference). The ICL8212 requires a voltage in excess of 1.15 volts to switch its output on (no current limit). Both devices have a low current output (HYSTERESIS) which is switched on for input voltages in excess of 1.15V. The HYSTERESIS output may be used to provide positive and noise free output switching using a simple feedback network.

Applications include:

1. Low voltage sensor/indicator
2. High voltage sensor/indicator
3. Non volatile out-of-voltage range sensor/indicator
4. Programmable voltage reference or zener diode
5. Series or shunt power supply regulator
6. Fixed value constant current source

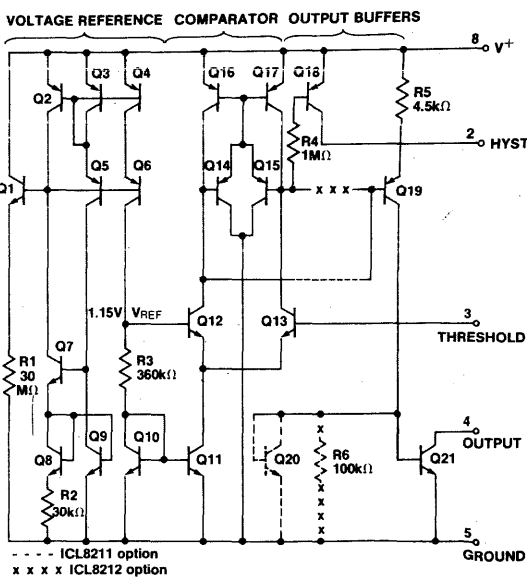
### PIN CONFIGURATION



### ORDERING INFORMATION

Part Number	Temperature Range	Package
ICL8211CPA	0 to +70°C	8 lead Mini DIP
ICL8211CTY	0 to +70°C	TO-99 Can
ICL8211MTY	-55° to +125°C	TO-99 Can
ICL8212CPA	0 to 70°C	8 lead Mini DIP
ICL8212CTY	0 to 70°C	TO-99 Can
ICL8212MTY	-55 to +125°C	TO-99 Can
ICL8211D	Dice only	
ICL8212D	Dice only	

### SCHEMATIC DIAGRAM



# ICL8211/ICL8212



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage .....	-0.5 to +30 volts
Output Voltage .....	-0.5 to +30 volts
Hysteresis Voltage .....	+0.5 to -10 volts
Threshold Input Voltage .....	+30 to -5 volts with respect to GROUND and +0 to -30 volts with respect to V <sup>+</sup>
Current into Any Terminal .....	±30mA
Power Dissipation (Note 1 & 2) .....	300mW
Operating Temperature Range ICL8211M/12M .....	-55°C to +125°C
Operating Temperature Range ICL8211C/12C .....	0 to +70°C
Storage Temperature Range .....	-65°C to +150°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**NOTE 1:** Rating applies for case temperatures to 125°C to ICL8211MTY/12MTY products. Derate linearly at -10mW/°C for ambient temperatures above 100°C.

**NOTE 2:** Derate linearly above 50°C by -10mW/°C for ICL8211C/12C products. The threshold input voltage may exceed +7 volts for short periods of time. However for continuous operation this voltage must be maintained at a value less than 7 volts.

## TYPICAL OPERATING CHARACTERISTICS (V<sup>+</sup> = 5V, T<sub>A</sub> = 25°C unless otherwise specified)

PARAMETER	SYMBOL	CONDITIONS	ICL8211			ICL8212			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
Supply Current	I <sup>+</sup>	2.0 < V <sup>+</sup> < 30 V <sub>T</sub> = 1.3V V <sub>T</sub> = 0.9V	10 50	22 140	40 250	50 10	110 20	250 40	μA μA
Threshold Trip Voltage	V <sub>TH</sub>	I <sub>OUT</sub> = 4mA V <sup>+</sup> = 5V V <sub>OUT</sub> = 2V V <sup>+</sup> = 2V V <sup>+</sup> = 30V	0.98 0.98 1.00	1.15 1.145 1.165	1.19 1.19 1.20	1.00 1.00 1.05	1.15 1.145 1.165	1.19 1.19 1.20	V V V
Threshold Voltage Disparity Between Output & Hysteresis Output	V <sub>THP</sub>	I <sub>OUT</sub> = 4 mA V <sub>OUT</sub> = 2V I <sub>HYST</sub> = 7μA V <sub>HYST</sub> = 3V		-8.0			-0.5		mV
Guaranteed Operating Supply Voltage Range	V <sub>SUPP</sub>	+25°C 0 to +70°C -55°C to +125°C	2.0 2.2 2.8		30 30 30	2.0 2.2 2.8		30 30 30	V V V
Typical Operating Supply Voltage Range	V <sub>SUPP</sub>	+25°C +125°C -55°C	1.8 1.4 2.5		30 30 30	1.8 1.4 2.5		30 30 30	V V V
Threshold Voltage Temperature Coefficient	ΔV <sub>TH</sub> /ΔT	I <sub>OUT</sub> = 4mA V <sub>OUT</sub> = 2V		+200			+200		ppm/°C
Variation of Threshold Voltage with Supply Voltage	ΔV <sub>TH</sub> /ΔV <sup>+</sup>	ΔV <sup>+</sup> = 10% at V <sup>+</sup> = 5V		1.0			1.0		mV
Threshold Input Current	I <sub>TH</sub>	V <sub>TH</sub> = 1.15V V <sub>TH</sub> = 1.00V		100 5	250		100 5	250	nA nA
Output Leakage Current	I <sub>OLK</sub>	V <sub>OUT</sub> = 30V V <sub>TH</sub> = 1.0V V <sub>OUT</sub> = 30V V <sub>TH</sub> = 1.3V V <sub>OUT</sub> = 5V V <sub>TH</sub> = 1.0V V <sub>OUT</sub> = 5V V <sub>TH</sub> = 1.3V			10 1			10 1	μA μA μA μA
Output Saturation Voltage	V <sub>SAT</sub>	I <sub>OUT</sub> = 4mA V <sub>TH</sub> = 1.0V V <sub>TH</sub> = 1.3V		0.17	0.4		0.17	0.4	V V
Max Available Output Current	I <sub>OH</sub>	(Note 3 & 4) V <sub>TH</sub> = 1.0V V <sub>OUT</sub> = 5V V <sub>TH</sub> = 1.3V -55°C ≤ T <sub>A</sub> ≤ 125°C V <sub>TH</sub> = 1.0V	4	7.0	12 15	15 12	35	mA mA	mA mA
Hysteresis Leakage Current	I <sub>LHYS</sub>	V <sup>+</sup> = 10V V <sub>TH</sub> = 1.0V V <sub>HYST</sub> = V <sup>-</sup>			0.1			0.1	μA
Hysteresis Sat Voltage	V <sub>HYS (max)</sub>	I <sub>HYST</sub> = -7μA V <sub>TH</sub> = 1.3V measured with respect to V <sup>+</sup>		-0.1	-0.2		-0.1	-0.2	V
Max Available Hysteresis Current	I <sub>HYS (max)</sub>	V <sub>TH</sub> = 1.3V	-15	-21		-15	-21		μA

**NOTE 3:** The maximum output current of the ICL8211 is limited by design to 15mA under any operating conditions. The output voltage may be sustained at any voltage up to +30 as long as the maximum power dissipation of the device is not exceeded.

**NOTE 4:** The maximum output current of the ICL8212 is not defined, and systems using the ICL8212 must therefore ensure that the output current does not exceed 30mA and that the maximum power dissipation of the device is not exceeded.



# ICM7206 CMOS Touch Tone™ Encoder

## FEATURES

- Low cost system with minimum component count
- Fully integrated oscillator uses 3.58 MHz color TV crystal
- High current bipolar output driver
- Low output harmonic distortion
- Wide operating supply voltage range: 3 to 6 volts
- Uses inexpensive single contact per key calculator type keyboard (ICM7206/C/D)
- Extremely low power  $\leq 5.5\text{mW}$  with a 5.5V supply
- Single and dual tone capabilities
- Multiple key lockout
- Disable output: provides output switch function whenever a key is pressed
- Custom options available

## GENERAL DESCRIPTION

The Intersil ICM7206/A/B/C/D are 2-of-8 sine wave tone encoders for use in telephone dialing systems. Each circuit contains a high frequency oscillator, two separate programmable dividers, a D/A converter, and a high level output driver.

The reference frequency is generated from a fully integrated oscillator requiring only a 3.58 MHz color TV crystal. This frequency is divided by 8 and is then gated into two divide by N counters (possible division ratios 1 through 128) which provide the correct division ratios for the upper and lower band of frequencies. The outputs from these two divide by N counters are further divided by 8 to provide the time sequencing for a 4 voltage level synthesis of each sinewave. Both sinewaves are added and buffered to a high current output driver, with provisions made for up to two external capacitors for low pass filtering, if desired. Typically, the total output harmonic distortion is 20% with no L.P. filtering and it may be reduced to typically less than 5% with filtering. The output drive level of the tone pairs will be approximately

-3dBV into a 900 ohm termination. The skew between the high and low groups is typically 2.5 dB without low pass filtering.

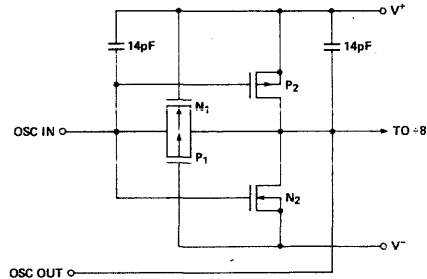
The 7206 uses either a 3 x 4 or 4 x 4 single contact keyboard; the oscillator will run whenever the power is applied, and the DISABLE output consists of a p-channel open drain FET whose source is connected to  $V^+$ .

The 7206A can also use a 3 x 4 or 4 x 4 keyboard, but requires a double contact type with the common line tied to  $V^+$ . The oscillator will be on whenever power is applied; the DISABLE output consists of a p-channel open drain FET; its' source is connected to  $V^+$ .

The 7206B requires a 4 x 4 double contact keyboard with the common line tied to  $V^-$ . The oscillator will be on only during the time that a ROW is enabled, and the DISABLE output consists of an n-channel open drain FET with its' source tied to  $V^-$ .

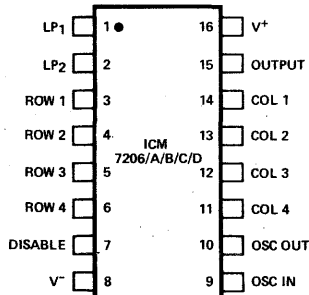
The 7206C uses either a 3 x 4 or 4 x 4 single contact keyboard; the oscillator will be on only during the time that a key is depressed. The DISABLE output consists of an n-channel open drain FET with its source tied to  $V^-$ .

The 7206D uses a single contact 3 x 4 or 4 x 4 keyboard. The oscillator will be on only during the time that a key is depressed. DISABLE output consists of a p-channel open drain FET with its source tied to  $V^+$ .



ICM7206 Oscillator

## PIN CONFIGURATION (OUTLINE DRAWING PE)



Pin 1 is designated either by a dot or a notch.

## ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
ICM7206 JPE	-40° C to +85° C	Plastic
ICM7206A JPE	-40° C to +85° C	Plastic
ICM7206B JPE	-40° C to +85° C	Plastic
ICM7206C JPE	-40° C to +85° C	Plastic
ICM7206D JPE	-40° C to +85° C	Plastic
ICM7206/D	-40° C to +85° C	DICE
ICM7206A/D	-40° C to +85° C	DICE
ICM7206B/D	-40° C to +85° C	DICE
ICM7206C/D	-40° C to +85° C	DICE
ICM7206D/D	-40° C to +85° C	DICE

# ICM7206 Family



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage (Note 2)	6.0V
Supply Current $V^-$ (terminal 8)	25mA
Supply Current $V^+$ (terminal 16)	40mA
Disable Output Volt. (term. 7)	Not more pos. than $V^+$ nor more neg. than $-6V$ with respect to $V^+$

Output Volt. (term. 15)	Not more pos. than $+5V$ with respect to $V^+$ , nor more neg. than $-1.0$ with respect to $V^-$
Output Current (terminal 15)	25mA
Power Dissipation	300mW
Operating Temperature Range	$-40^\circ C$ to $+85^\circ C$
Storage Temperature Range	$-55^\circ C$ to $+125^\circ C$

**NOTE 1.** Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**NOTE 2.** The ICM7206 family has a zener diode connected between  $V^+$  and  $V^-$  having a breakdown voltage between 6.2 and 7.0 volts. If the currents into terminals 8 and 16 are limited to 25 and 40mA maximum respectively, the supply voltage may be increased above 6 volts to zener voltage. With no such current limiting, the supply voltage must not exceed 6 volts.

## TYPICAL OPERATING CHARACTERISTICS

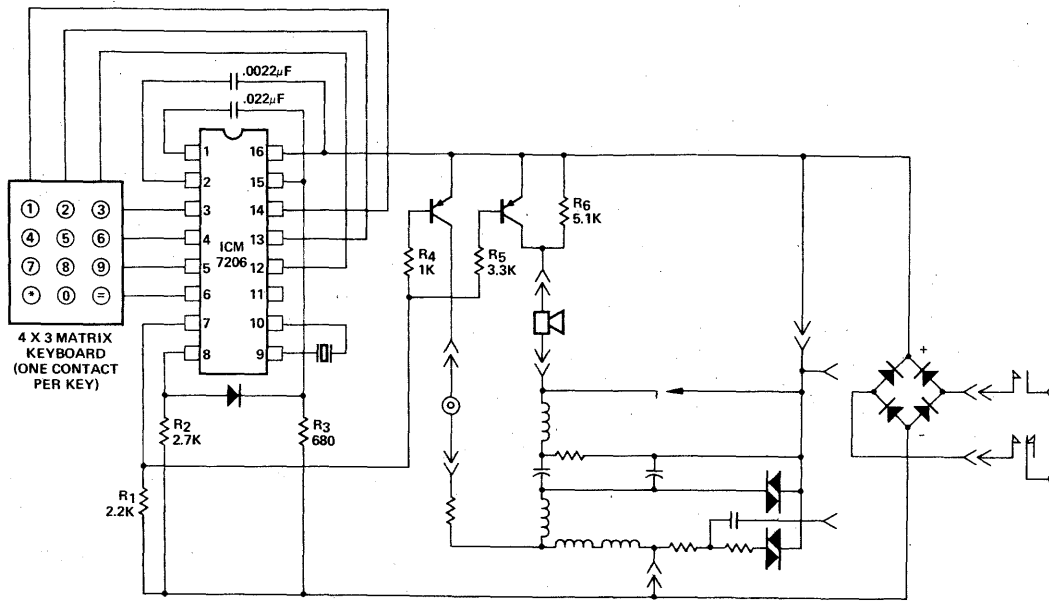
**TEST CONDITIONS:**  $V^+ = 5.5V$ , Test Circuit,  $T_A = 25^\circ C$  unless otherwise specified.

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS
Supply Current	$I^+$	$R_L$ disconnected		450	1000	$\mu A$
Guaranteed Operating Supply Voltage Range (Note 3)	$V_{OP}$	$-40^\circ C \leq T_A \leq +85^\circ C$	3.0		6.0	V
Peak to Peak Output Voltage	$V_{OUT}$	$C_1, C_2$ disconnected — Low Band	0.90	1.15	1.45	V
		$R_L = 1k\Omega$ , no filtering — High Band	1.10	1.40	1.70	
RMS Output Voltage	$R_L = 1k\Omega, f_{OUT} = 697Hz$	$C_2$ Only		480		mV
		$C_1$ to $C_2$		480		
		No filtering		490		
		$C_1$		490		
		$C_1$ to $C_2$		580		
		No filtering		655		
Skew Between High and Low Band Output Voltages		$R_L = 1k\Omega, C_1, C_2$ disconnected		2.5	3.0	dB
Output Impedance	$Z_O$	$R_L = 1k\Omega$	Operating	90	200	$\Omega$
			Quiescent	25		K $\Omega$
Total Output Harmonic Distortion	THD1	Either Hi or Low Bands No Low Pass Filtering		20	25	%
Total Output Harmonic Distortion	THD2	$R_L = 1k\Omega, C_1 = .002\mu F, f_{OUT} = 697Hz$		2.3	10	%
		$C_2 = 0.02\mu F, f_{OUT} = 1633Hz$		1.0	10	
Maximum Output Voltage Level	$V_{OH}$	$R_L = 1k\Omega$			4.6	V
Minimum Output Voltage Level	$V_{OL}$	$R_L = 1k\Omega$	0.5			
Keyboard Input Pullup Resistors	$R_{IN}$	Terminals 3,4,5,6,11,12,13,14	35	100	150	K $\Omega$
Keyboard Input Capacitance	$C_{IN}$	Terminals 3,4,5,6,11,12,13,14			5	pF
Guaranteed Oscillator Frequency Range (Note 4)	$f_{OSC}$	$3 \leq (V^+ - V^-) \leq 6V$	2.0		4.5	MHz
Guaranteed Oscillator Frequency Range		$4V \leq (V^+ - V^-) \leq 6V$	2.0		7	
System Startup Time on Application of Power	$t_{ON}$	ICM7206, ICM7206A		10		ms
System Startup Time on Application of Power and Key Depressed Simultaneously		ICM7206B, ICM7206C, ICM7206D			7	
DISABLE Output Saturation Resistance (ON STATE)	$R_D$	See Logic Table for Input Conditions Current = 4mA		330	700	$\Omega$
DISABLE Output Leakage (OFF STATE)	$I_{OLK}$	See Logic Table for Input Conditions			10	$\mu A$
Oscillator Load Capacitance	$C_{OSC}$	Measured between terminals 9 & 10, no supply voltage applied to circuit $-40^\circ C \leq T_A \leq 85^\circ C$		7		pF
Guaranteed Output Frequency Tolerance	$f_O$	Any output frequency Crystal tolerance $\pm 60ppm$ Crystal load capacitance $C_L = 30pF$			$\pm 0.75$	%
Oscillator Startup Time ICM7206B, C, D	$t_{START}$	$V^+ = 3V$ (Note 5)			7	ms

**NOTE 3:** Operation above 6 volts must employ supply current limiting. Refer to 'ABSOLUTE MAXIMUM RATINGS' and the Application Notes for further information.

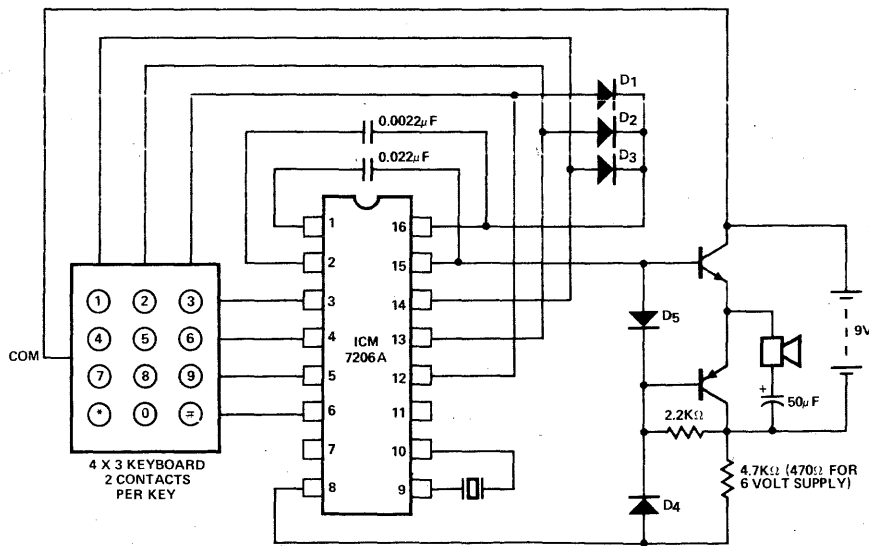
**NOTE 4:** The ICM7206 family uses dynamic high frequency circuitry in the initial 2<sup>3</sup> divider resulting in low power dissipation and excellent performance over a restricted frequency range. Thus, for reliable operation with a 6 volt supply an oscillator frequency of not less than 2MHz must be used.

**NOTE 5:** After row input is enabled.



**NOTE:** If dual contact keyboard is used, common should be left floating.

**FIGURE 5:** Telephone Handset Touch Tone Encoder



**FIGURE 6:** Portable Tone Generator

Intersil LINEAR





# ICM7555/7556 CMOS General Purpose Timers

## FEATURES

- Exact equivalent in most cases for SE/NE555/556 or the 355.
- Low Supply Current — 80 $\mu$ A Typ. (ICM7555)  
180 $\mu$ A Typ. (ICM7556)
- Extremely low trigger, threshold and reset currents - 20pA Typical
- High speed operation - 500 kHz guaranteed
- Wide operation supply voltage range guaranteed 2 to 18 volts
- Normal Reset function - No crowbaring of supply during output transition.
- Can be used with higher impedance timing elements than regular 555/6 for longer RC time constants.
- Timing from microseconds through hours
- Operates in both astable and monostable modes
- Adjustable duty cycle
- High output source/sink driver can drive TTL/CMOS
- Typical temperature stability of 0.005% per °C at 25°C
- Outputs have very low offsets, HI and LO

## GENERAL DESCRIPTION

The ICM7555/6 are CMOS RC timers providing significantly improved performance over the standard SE/NE555/6 and 355 timers, while at the same time being direct replacements for those devices in most applications. Improved parameters include low supply current, wide operating supply voltage range, low THRESHOLD, TRIGGER and RESET currents, no crowbaring of the supply current during output transitions, higher frequency performance and no requirement to decouple CONTROL VOLTAGE for stable operation.

Specifically, the ICM7555/6 are stable controllers capable of producing accurate time delays or frequencies. The ICM7555 is a dual ICM7555, with the two timers operating independently of each other, sharing only V<sup>+</sup> and GND. In the one shot mode, the pulse width of each circuit is precisely controlled by one external resistor and capacitor. For astable operation as an oscillator, the free running frequency and the duty cycle are both accurately controlled by two external resistors and one capacitor. Unlike the regular bipolar 555/6 devices, the CONTROL VOLTAGE terminal need not be decoupled with a capacitor. The circuits are triggered and reset on falling (negative) waveforms, and the output inverter can source or sink currents large enough to drive TTL loads, or provide minimal offsets to drive CMOS loads.

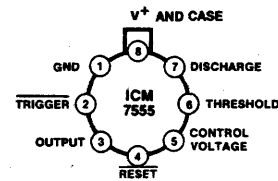
## APPLICATIONS

- Precision Timing
- Pulse Generation
- Sequential Timing
- Time Delay Generation
- Pulse Width Modulation
- Pulse Position Modulation
- Missing Pulse Detector

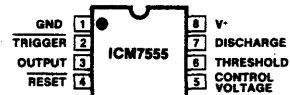
## ORDERING INFORMATION

ORDER PART NUMBER	TEMPERATURE RANGE	PACKAGE
ICM7555IPA	-20 to +85°C	8 Lead MiniDip
ICM7555ITV	-20 to +85°C	TO-99 Can
ICM7555MTV	-55 to +125°C*	TO-99 Can
ICM7556IPD	-20 to +85°C	14 Lead Plastic DIP
ICM7556MJD	-55 to +125°C*	14 Lead CERDIP
ICM7555/D		DICE
ICM7556/D		DICE

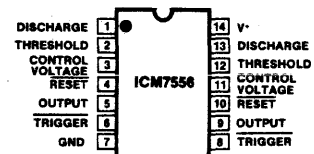
## PIN CONFIGURATIONS (Top View)



(OUTLINE DRAWING TV)



(OUTLINE DRAWING PA)



(OUTLINE DRAWING JD, PD)

\*Add /883B to order number if 883B processing is desired.

# ICM7555/ICM7556



## APPLICATION NOTES

### GENERAL

The ICM7555/6 devices are, in most instances, direct replacements for the NE/SE 555/6 devices. However, it is possible to effect economies in the external component count using the ICM7555/6. Because the bipolar 555/6 devices produce large crowbar currents in the output driver, it is necessary to decouple the power supply lines with a good capacitor close to the device. The 7555/6 devices produce no such transients. See Figure 2.

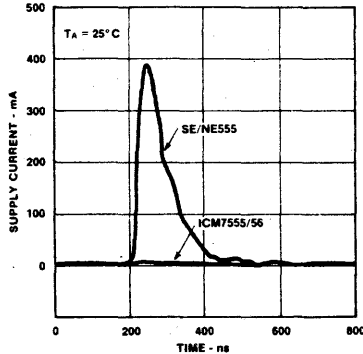


Figure 2. Supply Current Transient Compared with a Standard Bipolar 555 During an Output Transition

The ICM7555/6 produces supply current spikes of only 2-3 mA instead of 300-400 mA and supply decoupling is normally not necessary. Secondly, in most instances, the CONTROL VOLTAGE decoupling capacitors are not required since the input impedance of the CMOS comparators on chip are very high. Thus, for many applications 2 capacitors can be saved using an ICM7555, and 3 capacitors with an ICM7556.

### POWER SUPPLY CONSIDERATIONS

Although the supply current consumed by the ICM7555/6 devices is very low, the total system supply can be high unless the timing components are high impedance. Therefore, use high values for R and low values for C in Figures 3 and 4.

### OUTPUT DRIVE CAPABILITY

The output driver consists of a CMOS inverter capable of driving most logic families including CMOS and TTL. As such, if driving CMOS, the output swing at all supply voltages will equal the supply voltage. At a supply voltage of 4.5 volts or more the ICM7555/6 will drive at least 2 standard TTL loads.

### ASTABLE OPERATION

The circuit can be connected to trigger itself and free run as a multivibrator, see Figure 3. The output swings from rail to rail, and is a true 50% duty cycle square wave. (Trip points and output swings are symmetrical). Less than a 1% frequency variation is observed, over a voltage range of +5 to +15V.

$$f = \frac{1}{1.4 RC}$$

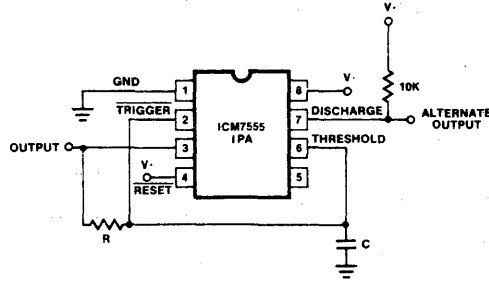


Figure 3: Astable Operation

### MONOSTABLE OPERATION

In this mode of operation, the timer functions as a one-shot. Initially the external capacitor (C) is held discharged by a transistor inside the timer. Upon application of a negative TRIGGER pulse to pin 2, the internal flip flop is set which releases the short circuit across the external capacitor and drives the OUTPUT high. The voltage across the capacitor now increases exponentially with a time constant  $t = R_A C$ . When the voltage across the capacitor equals  $2/3 V^+$ , the comparator resets the flip flop, which in turn discharges the capacitor rapidly and also drives the OUTPUT to its low state. TRIGGER must return to a high state before the OUTPUT can return to a low state.

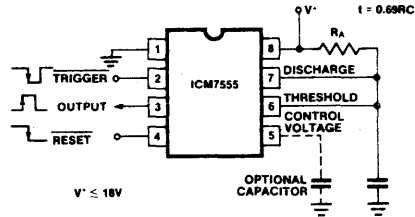


Figure 4: Monostable Operation

### CONTROL VOLTAGE

The CONTROL VOLTAGE terminal permits the two trip voltages for the THRESHOLD and TRIGGER internal comparators to be controlled. This provides the possibility of oscillation frequency modulation in the astable mode or even inhibition of oscillation, depending on the applied voltage. In the monostable mode, delay times can be changed by varying the applied voltage to the CONTROL VOLTAGE pin.

### RESET

The RESET terminal is designed to have essentially the same trip voltage as the standard bipolar 555/6, i.e. 0.6 to 0.7 volts. At all supply voltages it represents an extremely high input impedance. The mode of operation of the RESET function is, however, much improved over the standard bipolar 555/6 in that it controls only the internal flip flop, which in turn controls simultaneously the state of the OUTPUT and DISCHARGE pins. This avoids the multiple threshold problems sometimes encountered with slow falling edges in the bipolar devices.

Note: See other MPS parts in Interface and Military Sections.

## Precision Voltage References

- Second source to PMI
- Low temperature coefficient
- Guaranteed long term-stability
- Low noise

Model	Packages	Output Voltage	Regulating Current	Temp Coefficient (max, full temp)	Reverse Dynamic Impedance	Noise	Replaces
MP5010	All	1.22V	50 $\mu$ A – 500 $\mu$ A	50 ppm/C°	2 ohm	5 $\mu$ V	ICL8069 LM113 ZN423
MP5532/REF01	All	+10V	1.4 mA – 10 mA	8.5 ppm/C°	—	20 $\mu$ V P-P	REF01
MP5531/REF02	All	+5V	1.4 mA – 10 mA	8.5 ppm/C°	—	20 $\mu$ V P-P	REF02
MPREF05	All	+5V	1.4 mA – 10 mA	8.5 ppm/C°	—	20 $\mu$ V P-P	REF05
MPREF10	All	+10V	1.4 mA – 10 mA	8.5 ppm/C°	—	20 $\mu$ V P-P	REF10

## Precision Operational Amplifiers

- Second source to PMI
- Lowest noise in the industry
- Improved long term-stability
- High slew rate DAC follower

Model	Type	V <sub>OS</sub> (Max)	I <sub>B</sub> (Max)	Noise (nV/ $\sqrt{\text{Hz}}$ at 10 Hz)	Gain (Min)	CMRR (Min)	Slew Rate (typ)	Replaces
MP4136	Quad (741 Type)	6 mV	1.5 $\mu$ A		25 V/mV	70 dB	1.5 V/ $\mu$ sec	PM4136
MP5501/OP01	DAC Follower, Hi-speed	1 mV	50 nA		30 V/mV	85 dB	18 V/ $\mu$ sec	OP01
MP5502/OP02	Low Noise, Gen. Purpose	1.0 mV	60 nA	25 (typ)	50 V/mV	80 dB	25 V/ $\mu$ sec	OP02
MP5503/OP03	Dual, Matched	1.5 mV	60 nA	25 (typ)	50 V/mV	80 dB	.5 V/ $\mu$ sec	OP03
MP5504/OP04	Dual, Matched	1.5 mV	60 nA	25 (typ)	50 V/mV	80 dB	.5 V/ $\mu$ sec	OP04
MP5505/OP05	Low Noise, Instrumentation	.24 mV	$\pm$ 4 nA	18 (max)	200 V/mV	110 dB	.1 V/ $\mu$ sec	OP05
MP5507/OP07	Ultra Low Noise, Low V <sub>OS</sub> Instr.	60 $\mu$ V	4 nA	18 (max)	200 V/mV	106 dB	.1 V/ $\mu$ sec	OP07
MP5508/OP08	Low Input Current	.35 $\mu$ V	3 nA	22 (typ)	40 V/mV	100 dB	.12 V/ $\mu$ sec	OP08
MP5509/OP09	Quad Matched	1 mV	375 nA	18	50 V/mV	100 dB	1.0 V/ $\mu$ sec	OP09
MP5510/OP10	Dual 5505/OP05	.7 mV	$\pm$ 6 nA	18 (max)	150 V/mV	106 dB	.2 V/ $\mu$ sec	OP10
MP5511/OP11	Quad Matched	1 mV	375 nA	18	50 V/mV	100 dB	1.0 V/ $\mu$ sec	OP11
MP5512/OP12	Low Input Current	.26 mV	2.6 nA	22 (typ)	25 V/mV	100 dB	.12 V/ $\mu$ sec	OP12
MP5514/OP14	Dual Matched High Performance	1.5 mV	60 nA	25	50 V/mV	80 dB	.5 V/ $\mu$ sec	OP14
MP5527/OP27	Ultra Low Noise	60 $\mu$ V	50 nA	5.5 (max)	600 V/mV	108 dB	2.8 V/ $\mu$ sec	OP27
MP5537/OP37	Ultra Low Noise, High Speed	60 $\mu$ V	50 nA	5.5 (max)	600 V/mV	108 dB	17 V/ $\mu$ sec	OP37
MP207/OP207	Dual, Matched OP07	230 $\mu$ V	5.6 nA	18.0 (max)	150 V/mV	130 dB	0.2 V/ $\mu$ sec	OP207
MP227/OP227	Ultra Low Noise, Low V <sub>OS</sub> Instr.	180 $\mu$ V	$\pm$ 60 nA	6.0 (max)	600 V/mV	108 dB	2.8 V/ $\mu$ sec	OP227
MPLM108/208/308	Super Beta	1.0 mV	3 nA		40 V/mV	96 dB		LM108/208/308
MPOP2108/2208/2308	Low Input Current	1.0 mV	3 nA		40 V/mV	96 dB		OP2108/2208/2308

Note: All specifications: -55°C to +125°C

All Micro Power Systems' IC's are available with 883B processing, in dice form, or to customer's specifications.



# MOTOROLA SEMICONDUCTORS

P.O. BOX 20912 • PHOENIX, ARIZONA 85036

## OPERATIONAL AMPLIFIERS

### HIGH SLEW RATE, WIDE BANDWIDTH, SINGLE SUPPLY QUAD OPERATIONAL AMPLIFIER

A standard low-cost Bipolar technology with innovative design concepts is employed for the MC34074 series of monolithic quad operational amplifiers. These devices offer 4.5 MHz of gain bandwidth product, 13 V/ $\mu$ s slew rate, and fast settling time without the use of JFET device technology. In addition, low input offset voltage can economically be achieved. Although these devices can be operated from split supplies, they are particularly suited for single supply operation, since the common mode input voltage range includes ground potential ( $V_{EE}$ ). The all NPN output stage, characterized by no deadband crossover distortion and large output voltage swing, also provides high capacitive drive capability, excellent phase and gain margins, low open-loop high frequency output impedance and symmetrical source/sink ac frequency response.

The MC34074/33074/35074 series of devices are available in standard or prime performance (A Suffix) grades and specified over commercial, industrial/vehicular or military temperature ranges.

#### Features

- Wide Bandwidth: 4.5 MHz
- High Slew Rate: 13 V/ $\mu$ s
- Fast Settling Time: 1.1  $\mu$ s to 0.10%
- Wide Single Supply Operating Range: 3.0 to 44 Volts
- Wide Input Common Mode Range Including Ground ( $V_{EE}$ )
- Low Input Offset Voltage: 2.0 mV Maximum (A Suffix)
- Large Output Voltage Swing: -14.7 V to +14.0 V for  $V_S = \pm 15$  V
- Large Capacitance Drive Capability: 0 to 10,000 pF
- Low T.H.D. Distortion: 0.02%
- Excellent Phase Margins: 60°

### GENERAL OP-AMP SELECTOR GUIDE

#### General Purpose Op-Amps

	Single	Dual	Quad
Compensated	LF351 LF355 LF356 LF357 MC1741 MC34001 TL081	LF353 LM358 MC1458 MC1747 MC3458 MC4558 MC34002 TL082	LF347 LM324 MC3403 MC4741 MC34004 TL084
Uncompensated	LM301A MC1709 MC1748	MC1437	

#### JFET (BIFET) INPUT — High Impedance

Single	Dual	Quad
LF351 LF355 LF356 LF357 MC34001 TL071 TL081	LF353 MC34002 TL072 TL082	LF347 MC34004 TL074 TL084

#### High Slew Rate

Slew Rate (V/ $\mu$ s)			
5.0	8.0 to 20		50
LF355	LF347 LF351 LF353	MC34001 MC34002 MC34004 OP-37	LF357
	LF356 MC1458S MC1741S MC34074	TL071 TL072 TL074 TL081 TL082 TL084	

#### Programmable

Adjustable bias current varies noise bandwidth, power consumption, etc.

Single
MC1776 MC3476

#### Low Noise

Single	Dual	Quad
MC1741N OP-27/OP-37 TL071	MC1458N TL072	TL074

#### Wide Bandwidth

Bandwidth (MHz)				
3.0	4.0	5.0	20	100
MC4558S	MC34001 MC34002 MC34004	MC3401 LF356 MC34074	LF357	NE592 MC1733

#### Low Offset Voltage

$V_{IO}$ Max ( $\mu$ V)			
$\leq 200$	500	2000	
OP-27/OP-37	LM11C LM308A	MC34001A MC34002A MC34074A	MC1709A LM301A LM308

#### Low Drift Over Temperature ( $TCV_{IO}$ )

$TCV_{IO} \leq 5.0 \mu V/^{\circ}C$ Max*	
LM11	LM308A OP-27/OP-37

\*Input offset voltage temperature coefficient

#### Single-Supply

Dual	Quad
LM358 LM2904 MC3458 MC3405	LM324 LM2900 LM2902 LM3900 MC3301 MC3401 MC3403 MC34074



**MOTOROLA**

# Voltage Regulators

## Fixed Output Voltage Regulators

- Low-cost monolithic circuits for positive and/or negative regulation at currents from 100 mA to 3.0 A
- Ideal for on-card regulation of subsystems
- Internal current limiting thermal shutdown and safe-area compensation

## Fixed/Voltage, 3-Terminal Regulators for Positive or Negative Polarity Power Supplies

V <sub>out</sub> Volts	Tol.† Volts	I <sub>o</sub> mA Max	Device Positive Output	Device Negative Output
2	±0.1	1500	—	MC7902C
3	±0.15	100	—	MC79L03AC
	±0.3		—	MC79L03C
5	±0.5	100	MC78L05C	MC79L05C
			MC78L05AC	MC79L05AC
	±0.25	500	MC78M05C	MC79M05C
			—	—
	±0.4	1500	LM109	—
			LM209	—
	±0.25	—	LM309	—
	±0.35	—	MC7805*	—
	±0.25	—	MC7805B#	—
			MC7805C	MC7905C
	±0.2	—	MC7805A*	—
			MC7805AC	MC7905AC
	±0.25	—	LM140-5*	—
	±0.2	—	LM140A-5*	—
	±0.25	—	LM340-5	—
	±0.2	—	LM340A-5	—
	±0.1	—	TL780-05C	—
	±0.25	3000	MC78T05*	—
			MC78T05C	—
	±0.2	—	MC78T05A*	—
MC78T05AC			—	
±0.4	—	LM123*	—	
		LM223	—	
±0.25	—	LM323	—	
±0.2	—	LM123A	—	
		LM223A	—	
—	—	LM323A	—	
5.2	±0.26	1500	—	MC7905.2C
6	±0.3	500	MC78M06C	—
			—	—
	±0.35	1500	MC7806*	—
			MC7806B#	—
	±0.3	—	MC7806C	MC7906C
			MC7806A*	—
	±0.24	—	MC7806AC	—
			LM140-6*	—
±0.3	—	LM340-6	—	
		3000	MC78T06*	—
—	—	—	MC78T06C	—

(continued)

V <sub>out</sub> Volts	Tol.† Volts	I <sub>o</sub> mA Max	Device Positive Output	Device Negative Output
8	±0.8	100	MC78L08C	—
			MC78L08AC	—
	±0.4	500	MC78M08C	—
			MC7808*	—
		1500	MC7808B#	—
			MC7808C	MC7908C
	±0.3	—	MC7808A*	—
			MC7808AC	—
	±0.4	—	LM140-8*	—
			LM340-8	—
—	—	3000	MC78T08*	—
		—	MC78T08C	—
12	±1.2	100	MC78L12C	MC79L12C
			MC78L12AC	MC79L12AC
	±0.6	500	MC78M12C	MC79M12C
			—	—
	±0.5	1500	MC7812*	—
			MC7812B#	—
			MC7812C	MC7912C
			MC7812A*	—
	±0.6	—	MC7812AC	—
			LM140-12*	—
	±0.5	—	LM140A-12*	—
	±0.6	—	LM340-12	—
	±0.5	—	LM340A-12	—
	±0.24	—	TL780-12C	—
±0.6	3000	MC78T12*	—	
		MC78T12C	—	
		MC78T12A*	—	
		MC78T12AC	—	
15	±1.5	100	MC78L15C	MC78L15C
			MC78L15AC	MC78L15A
	±0.75	500	MC78M15C	MC79M15C
			—	—
	±0.6	1500	MC7815*	—
			MC7815B#	—
			MC7815C	MC7915C
			MC7815A*	—
	±0.75	—	MC7815AC	—
			LM140-15*	—
±0.6	—	LM140A-15*	—	
±0.75	—	LM340-15	—	
±0.6	—	LM340A-15	—	
±0.3	—	TL780-15C	—	
±0.75	3000	MC78T15*	—	
		MC78T15C	—	
		MC78T15A*	—	
±0.6	—	—	MC78T15AC	—

#T<sub>J</sub> = -40 to +125°C

\*T<sub>J</sub> = -55 to +150°C

†Output Voltage Tolerance for Worst Case

**Fixed Output Voltage Regulators (continued)**

V <sub>out</sub> Volts	Tol.† Volts	I <sub>o</sub> mA Max	Device	
			Positive Output	Negative Output
18	±1.8	100	MC78L18C	MC79L18C
			MC78L18AC	MC79L18AC
		500	MC78M18C	—
	±0.7	1500	MC7818*	—
			MC7818B#	—
			MC7818C	MC7918C
			MC7818A*	—
MC7818AC	—			
±0.9	3000	LM140-18*	—	
		LM340-18	—	
		MC78T18*	—	
		MC78T18C	—	
20	±1.0	500	MC78M20C	—
24	±2.4	100	MC78L24C	MC79L24C
			MC78L24AC	MC79L24AC
		500	MC78M24C	—
	±1.0	1500	MC7824*	—
			MC7824B#	—
			MC7824C	MC7924C
			MC7824A*	—
	MC7824AC	—		
	±1.2	3000	LM140-24*	—
			LM340-24	—
MC78T24*			—	
MC78T24C	—			

#T<sub>J</sub> = -40 to +125°C †Output Voltage Tolerance for Worst Case  
\*T<sub>J</sub> = -55 to +150°C

**Positive Output Regulators (continued)**

I <sub>o</sub> mA Max	Device	S u f f i x	V <sub>out</sub> Volts		V <sub>in</sub> Volts		V <sub>in</sub> — V <sub>out</sub> Differ- ential Volts Min
			Min	Max	Min	Max	
500	LM317M	T	1.2	37	5.0	40	3.0
	LM317M	R					
	LM217M#						
	LM117M*						
600	MC1469	R	2.5	32	9.0	35	3.0
	MC1569			37		40	
1500	LM317	T	1.2	37	5.0	40	3.0
	LM317	H, K					
	LM217#						
	LM117*						
3000	LM350	T	1.2	33	5.0	36	3.0
	LM350	K					
	LM250#						
	LM150*						

#T<sub>J</sub> = -25 to +150°C  
\*T<sub>J</sub> = -55 to +150°C

**Adjustable Output Voltage Regulators**
**Positive Output Regulators**

I <sub>o</sub> mA Max	Device	S u f f i x	V <sub>out</sub> Volts		V <sub>in</sub> Volts		V <sub>in</sub> — V <sub>out</sub> Differ- ential Volts Min
			Min	Max	Min	Max	
100	LM317L	H, Z	1.2	37	5.0	40	3.0
	LM217L#						
	LM117L*						
	MC1723						
CG							
G							
CL							
L							
CD							
250	MC1469	G	2.5	32	9.0	35	3.0
	MC1569			37		40	

#T<sub>J</sub> = -40 to +125°C  
\*T<sub>J</sub> = -55 to +150°C  
†Output Voltage Tolerance for Worst Case

**Negative Output Regulators**

I <sub>o</sub> mA Max	Device	S u f f i x	V <sub>out</sub> Volts		V <sub>in</sub> Volts		V <sub>in</sub> — V <sub>out</sub> Differ- ential Volts Min
			Min	Max	Min	Max	
250	MC1463	G	-3.8	-32	9.0	35	3.0
	MC1563		-3.6	-33			
500	LM337M	T	-1.2	-37	5.0	40	3.0
	LM337M	R					
	LM237M#						
	LM137M*						
600	MC1463	R	-3.8	-34	9.0	35	3.0
	MC1563						
1500	LM337	T	-1.2	-37	5.0	40	3.0
	LM337	H, K					
	LM237#						
	LM137*						

#T<sub>J</sub> = -25 to +150°C  
\*T<sub>J</sub> = -55 to +150°C



**Switching Regulators**

Used as the control circuit in PWM, push-pull, bridge and series type switchmode supplies. The devices include the reference, oscillator, pulse-width modulator, phase splitter and output sections. Frequency and duty cycle are independently adjustable.

I <sub>O</sub> mA Max	V <sub>CC</sub> Volts		f <sub>o</sub> kHz		Device	Suffix	T <sub>A</sub> °C	Case
	Min	Max	Min	Max				
40	1	30	2.0	100	MC3420	P	0 to +70	648
						L		620
					MC3520	L	-55 to +125	620
250*	7.0	40	1.0	300	MC34060	P	0 to +70	646
						L		632
					MC35060	L	-55 to +125	632
250	7.0	40	1.0	300	TL494	CN	0 to +70	648
						CJ		620
						IN	-25 to +85	648
						IJ		620
						MJ		-55 to +125
250		>40	1.0	300	TL495**	CN	0 to +70	707
						CJ		726
						IN	-25 to +85	707
						IJ	-25 to +85	726
±400	8	40	0.1	400	SG3525A SG3525A SG2525A SG2525A SG1525A	N	0° to +70	648
						J	0 to +70	620
						N	-40 to +85	648
						J		620
						J	-55 to +125	620
±400	8	40	0.1	400	SG3527A SG3527A SG2527A SG2527A SG1527A	N	0 to +70	648
						J		620
						N	-40 to +85	648
						J		620
						J	-55 to +125	620
±200	8	40	0.001	400	SG3526 SG3526 SG2526 SG2526 SG1526	N	0 to +70	707
						J		726
						N	-40 to +85	707
						J		726
						J	-55 to +125	726
1500*	2.5	40	0.1	100	μA78S40 μA78S40 μA78S40	PC	0 to +70	648
						DC		620
						DM	-55 to +125	
1500*	2.5	40	0.1	100	MC34063 MC34063 MC33063 MC33063 MC35063	PI	0 to +70	626
						U		693
						PI	-40 to +85	626
						U		693
						U		-55 to +125

\*Single output device  
\*\*Internal 39 V zener for <40 volt operation

**LINEAR**

Motorola Semiconductor



**MOTOROLA**

# SEMICONDUCTORS

P.O. BOX 20912 • PHOENIX, ARIZONA 85036

**PRECISION REFERENCES**  
**MC1400 MC1400A**  
**MC1500 MC1500A**

## Tight-Tolerance, Low-Drift Voltage Reference Family

The MC1400 series of ICs is a family of temperature-compensated voltage references for precision data conversion and instrumentation applications. Advances in thin-film resistors, laser-trimming techniques, ion-implanted devices, and monolithic fabrication techniques make this reference both temperature and time stable in applications demanding accuracy to the 12-bit level.

## PRECISION VOLTAGE REFERENCES

2.5, 5.0, 6.25 and 10-VOLT

**LASER-TRIMMED SILICON MONOLITHIC INTEGRATED CIRCUIT**

### Features

- Four Different Output Voltages: 2.5, 5.0, 6.25, 10 V
- Tight Absolute Accuracy:  $\pm 0.2\%$  Maximum Initial Tolerance
- Single-Component Output Trimming Without Degrading Temperature Coefficient
- Wide Input Voltage Range:  $(V_{out} + 1.0 V) \leq V_{in} \leq 40 V$
- Three-Terminal Operation: Positive References That Can Source and Sink Current
- Two-Terminal Operation: Positive or Negative References Floating References
- Low Current Consumption: 1.0 mA Typical
- Very Low Temperature Coefficient
- Low Output Noise Voltage
- Excellent Ripple Rejection: 87 dB Typical at 120 Hz
- Excellent Long Term Stability: 25 ppm/1000 Hrs Typical

## Special Regulators

### Floating Voltage and Current Regulators

Designed for laboratory type power supplies. Voltage is limited only by the break down voltage of associated, external, series-pass transistors.

V <sub>out</sub> Volts		I <sub>O</sub> mA Max	Device	S u f f i x	V <sub>aux</sub> Volts		P <sub>D</sub> Watts Max	$\Delta V_{ref}/V_{ref}$ %		$\Delta I_L/L$ % Max	TC V <sub>out</sub> %/°C Typ	Case
Min	Max				Min	Max		Line	Load			
0	*	*	MC1466	L	21	30	0.75	0.015	0.015	0.2	0.001	632
			MC1566	L	20	35		0.004	0.004		0.1	

\*Dependent on characteristics of external series-pass elements.

### Dual $\pm 15$ V Tracking Regulators

Internally, the device is set for  $\pm 15$  V, but an external adjustment can change both outputs simultaneously, from 8.0 V to 20 V.

V <sub>out</sub> Volts		I <sub>O</sub> mA Max	V <sub>in</sub> Volts		Device	S u f f i x	P <sub>D</sub> Watts Max	Regline mV	Regload mV	TC %/°C (T <sub>low</sub> to T <sub>high</sub> ) Typ	T <sub>A</sub> °C	Case	
Min	Max		Min	Max									
14.8	15.2	$\pm 100$	17	30	MC1468	G	0.8	10	10	3.0	0 to +75	603C	
						L	1.0					632	
						R	2.4					614	
					MC1568	G	0.8					-55 to +125	603C
						L	1.0						632
						R	2.4						614

(continued)

LINEAR  
Motorola Semiconductor





**Special Regulators (continued)**

**Low Temperature Drift, Voltage References**

V <sub>out</sub> Volts Typ	I <sub>O</sub> mA Max	ΔV <sub>out</sub> /ΔT ppm/°C Max	Device	Regline mV Max	Regload mV Max	T <sub>A</sub> °C	Case
1.235 ± 12 mV	20	20 Typ	LM385BZ-1.2	(Note 1)	1.0 (Note 2)	0 to +70	29
			LM285Z-1.2			-40 to +85	
1.235 ± 25 mV			LM385Z-1.2			0 to +70	
2.5 ± 38 mV			LM385BZ-2.5		2.0 (Note 3)	-40 to +85	
			LM285Z-2.5			0 to +70	
2.5 ± 75 mV			LM385Z-2.5				
2.5 ± 5.0 mV	±10	25	MC1400G2	3.0  (Note 4)	10  (Note 7)	0 to +70	601
		10	MC1400AG2				
		40	MC1500G2				
		10	MC1500AG2				
2.5 ± 25 mV	10	40	MC1403	3.0/4.5  (Note 5)	10  (Note 8)	0 to +70	693, 79, 751
		25	MC1403A				693, 79
		55	MC1503				-55 to +125
		25	MC1503A				
5.0 ± 10 mV	±10	25	MC1400G5	4.0  (Note 4)	20  (Note 7)	0 to +70	693
		10	MC1400AG5				
		40	MC1500G5				
		10	MC1500AG5				
5.0 ± 50 mV	10	40	MC1404U5	6.0  (Note 6)	10  (Note 8)	0 to +70	
		25	MC1404AU5				
		55	MC1504U5				
		25	MC1504AU5				
6.25 ± 10 mV	±10	25	MC1400G6	4.0  (Note 4)	20  (Note 7)	0 to +70	601
		10	MC1400AG6				
		40	MC1500G6				
		10	MC1500AG6				
6.25 ± 60 mV	10	40	MC1404U6	6.0  (Note 6)	10  (Note 8)	0 to +70	693
		25	MC1404AU6				
		55	MC1504U6				
		25	MC1504AU6				
10 ± 20 mV	±10	25	MC1400G10	4.0  (Note 4)	20  (Note 7)	0 to +70	601
		10	MC1400AG10				
		40	MC1500G10				
		10	MC1500AG10				
10 ± 100 mV	10	40	MC1404U10	6.0  (Note 6)	10  (Note 8)	0 to +70	693
		25	MC1404AU10				
		55	MC1504U10				
		25	MC1504AU10				
2.5 to 37	100	50 Typ	TL431C	Shunt Reference Dynamic Impedance z ≤ 0.5 Ω		0 to +70	29, 626
			TL431I			-40 to +85	693
			TL431M			-55 to +125	693

**Notes:**

1. Micro-Power Reference Diode Dynamic Impedance (z) ≤ 1.0 Ω at I<sub>R</sub> = 100 μA
2. 10 μA ≤ I<sub>R</sub> ≤ 1.0 mA
3. 20 μA ≤ I<sub>R</sub> ≤ 1.0 mA
4. (V<sub>out</sub> + 1.0 V) ≤ V<sub>in</sub> ≤ 40 V
5. 4.5 V ≤ V<sub>in</sub> ≤ 15 V/15 V ≤ V<sub>in</sub> ≤ 40 V
6. (V<sub>out</sub> + 2.5 V) ≤ V<sub>in</sub> ≤ 40 V
7. -10 mA ≤ I<sub>L</sub> ≤ +10 mA
8. 0 mA ≤ I<sub>L</sub> ≤ 10 mA



# MOTOROLA SEMICONDUCTORS

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## MOS Telecommunications Products

### MC14400/01/02/03/05 Mono-Circuit

- On-board Reference
- A-Law/Mu-Law Companding and D3/D4/CCITT Format
- Synchronous and Asynchronous Operation
- 16/18/22 Pin Packages

### MC14407

- Per-Channel Full-Duplex Capability
- Power-Down Input
- Mu-Law Digital Format (MC14406/07)
- A-Law CCITT Digital Format (MC14404)

### MC14408/09 Binary to Pulse Dialer

- On-Chip Oscillator
- Dialing of Numbers Up to 16 Digits
- Memory Storage and Re-Dialing of Last Telephone Number

### MC14410 Tone Encoder

- Single-Tone Capability
- Fast Oscillator Turn-On and Turn-Off Times

### MC14411 Baud Rate Generator

- Sixteen Different Output Clock Rates
- Programmable Time Bases for One-of-Four Multiple Output Rates
- Buffered Outputs Compatible with Low-Power TTL

### MC14412 0-600 Baud Modem

- Originate and Answer Mode
- Simplex, Half-Duplex and Full-Duplex Operation

### MC14413/14 and 145414 PCM Low Pass/Bandpass Filters

- Transmit Bandpass and Received Low Pass (MC14413)
- Transmit and Receive Low Pass (MC14414/5414)
- Single-Supply Capability
- $\pm 5$  to  $\pm 8$  Volt Power Supply Ranges
- Two Operational Amplifiers Available

### MC14416/17/18 Time Slot Assigner Circuit

- Independent Transmit and Receive Frame Syncs and Enables
- Up to 64 Times Slots Per Frame
- TTL and CMOS Level Compatibility

### MC14419 2-of-8 Tone Encoder

- Suppressed Output for Illegal Input Codes
- Codes for Numbers 0-9 Produce a Strobe Pulse

### MC142100/5100 Crosspoint Switch

- Power-On Reset (MC145100 Only)
- Large Analog Range

### MC145415 Linear Phase Filter

- Dual Low Pass Linear Phase Filter
- $\pm 5$  to  $\pm 8$  Volt Power Supply Range
- Two Op Amps Available

### MC145422 2-Wire UDLT

- 2-Wire Master Universal Data Loop Transceiver
- For Voice/Data Transceiver Applications
- Data Rate of 80 kbps

### MC145426 2-Wire UDLT

- 2-Wire Slave Universal Data Loop Transceiver
- For Voice Data Transceiver Applications
- Data Rate of 80 kbps

### MC145428 Data Set Interface Chip

- Provides Synchronous to Asynchronous Data Conversion
- Provides Asynchronous to Synchronous Data Conversion
- Synchronous Data Rates of 2.048 Mbps
- Asynchronous Data Rates of 128 kbps

### MC145429 Telsel Audio Interface Chip

- Provides CPU Control of an Analog Signal for Earpiece, Mouthpiece and Speaker Phone
- Provides CPU Control of Analog Signal Gain and Attenuation
- Loop Back Test Feature

### MC145432 2600 Hz Tone Signaling Filter

- 2600 Hz Notch and Bandpass Filter
- 2600 Hz Sinewave Generator
- Uncommitted Op Amp Capable of Driving a 600 ohm Load
- On Board Crystal Oscillator or External Clock

### MC145433 Tuneable Notch/Bandpass Filter

- Independently Tuneable Notch and Bandpass Filter
- Clock Output Pin
- Includes an Uncommitted Op Amp Capable of Driving 600 Loads
- TTL or CMOS Compatible Digital Inputs
- On Board Crystal Oscillator or External Clocks

### MC145440/441 300 Baud Modem Filter (Bell 103/CCITT)

- Answer or Originate Mode
- Low Band Bandpass Filter
- High Band Bandpass Filter

### MC145445 0-600 Baud Modem

- Originate and Answer Mode
- Bell 103/113 and CCITT V.21

### MC145450 0-1200 Baud Modem

- Bell 202 and CCITT V.23
- On Board Crystal Oscillator or External Clock

### MC6860 0-600 bps Digital Modem

- Originate and Answer Modes
- Crystal or External Reference Control
- Full-Duplex or Half-Duplex Operation
- Automatic Answer and Disconnect

### MC6172 2400 bps Digital Modulator (Formerly 6862)

- 511-Bit CCITT Test Pattern
- CCITT and U.S. Phase Options (Bell 201)
- 1200/2400 bps operation
- Answer-Back Tone

### MC6173 2400 bps Demodulator

- Compatible with MC6172 Modulator
- CCITT and U.S. Phase Options
- Compatible Functions for 201B/C Data Sets

# ADC0844 8-Bit $\mu$ P Compatible A/D Converter with 4-Channel Multiplexer

## General Description

The ADC0844 is a CMOS 8-bit successive approximation A/D converter with a versatile analog input multiplexer. The 4-channel multiplexer can be software configured for single-ended, differential or pseudo-differential modes of operation.

The differential mode provides low frequency input common-mode rejection and allows offsetting the analog range of the converter. In addition, the A/D's reference can be adjusted enabling the conversion of reduced analog ranges with 8-bit resolution.

This A/D is designed to operate from the control bus of the NSC800™ and the wide variety of 8080  $\mu$ P derivatives. TRI-STATE® output latches that directly drive the data bus permit this A/D to be configured as a memory location or as an I/O device to the microprocessor with no interface logic necessary.

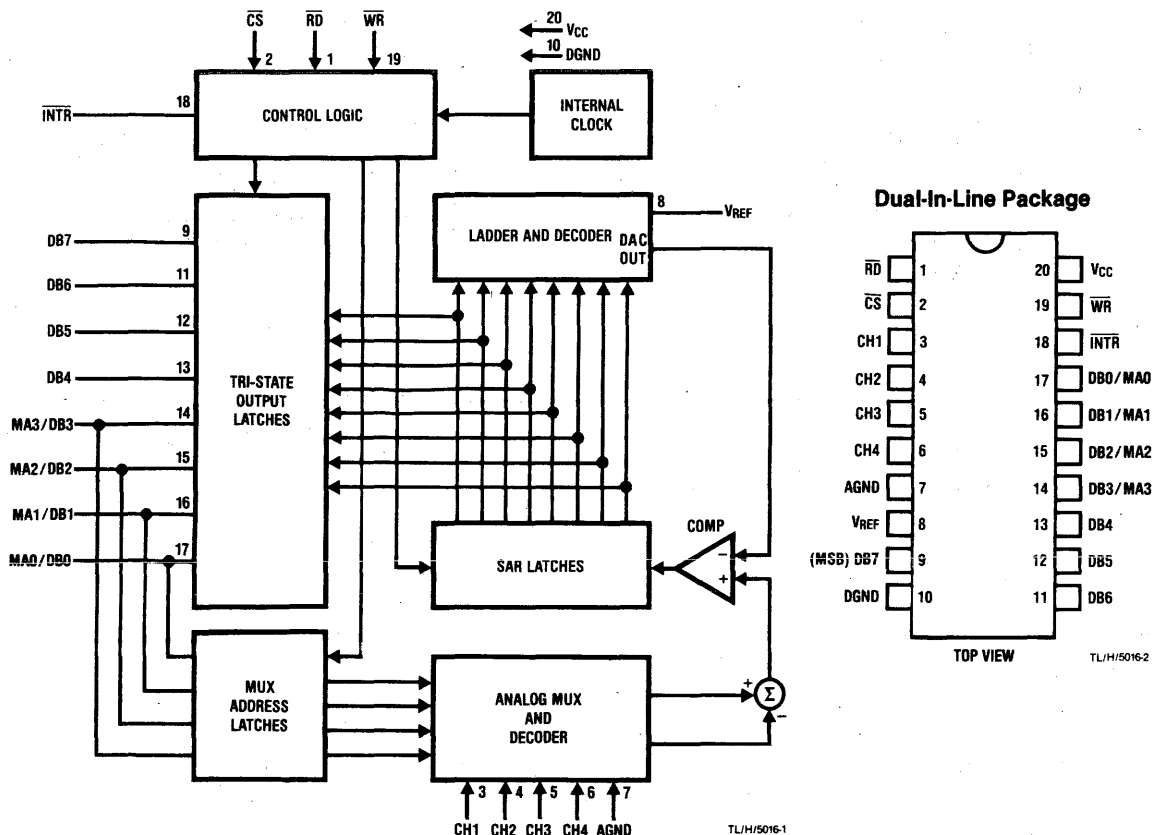
## Features

- Compatible with 8080  $\mu$ P derivatives—no interface logic needed
- Easy interface to all microprocessors
- Operates ratiometrically or with 5V<sub>DC</sub> voltage reference
- No zero or full-scale adjust required
- 4-channel multiplexer with address logic
- Internal clock
- 0V to 5V input range with single 5V power supply
- T<sup>2</sup>L/MOS input/output compatible
- 0.3" standard width 20-pin DIP

## Key Specifications

- |                          |                               |
|--------------------------|-------------------------------|
| ■ Resolution             | 8 Bits                        |
| ■ Total Unadjusted Error | $\pm 1/2$ LSB and $\pm 1$ LSB |
| ■ Single Supply          | 5V <sub>DC</sub>              |
| ■ Low Power              | 10mW                          |
| ■ Conversion Time        | 40 $\mu$ s                    |

## Block and Connection Diagrams



NSC800™ is a trademark of National Semiconductor Corp.

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## LF13006, LF13007 Digital Gain Set

### General Description

The LF13006, LF13007 are precision digital gain sets used for accurately setting non-inverting op amp gains. Gains are set with a 3-bit digital word which can be latched in with  $\overline{WR}$  and  $\overline{CS}$  pins. All digital inputs are TTL and CMOS compatible.

The LF13006 shown below will set binary scaled gains of 1, 2, 4, 8, 16, 32, 64, and 128. The LF13007 will set gains of 1, 2, 5, 10, 20, 50, and 100 (a common attenuator sequence). In addition, both versions have several taps and two uncommitted matching resistors which allow customization of the gain.

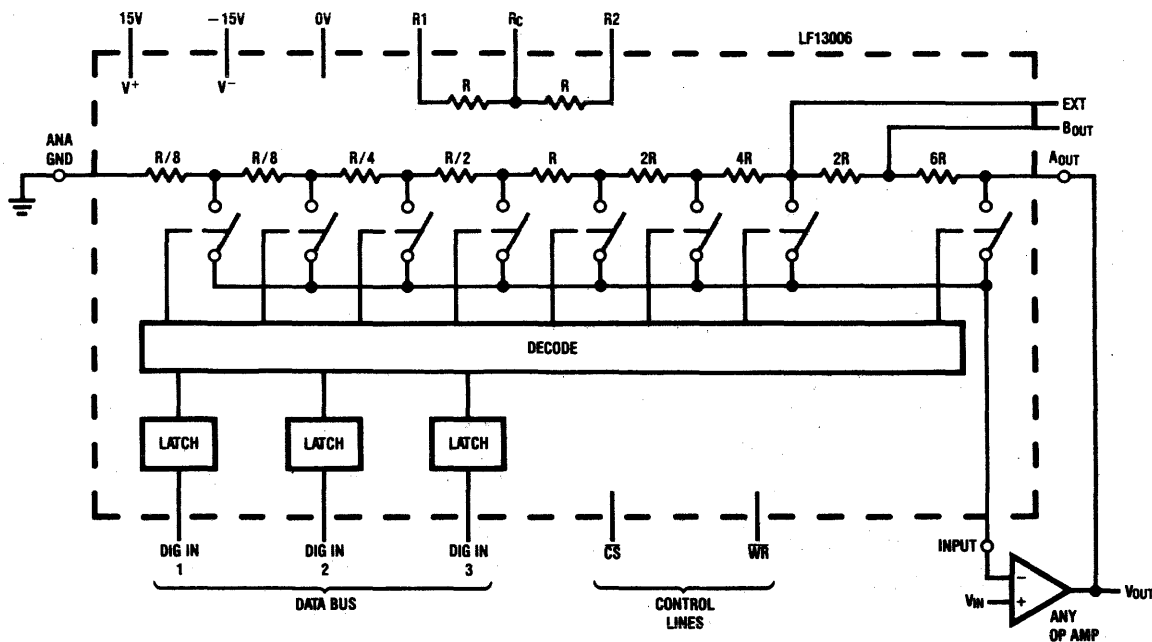
The gains are set with precision thin film resistors. The low temperature coefficient of the thin film resistors and their excellent tracking result in gain ratios which are virtually independent of temperature.

The LF13006, LF13007 used in conjunction with an amplifier not only satisfies the need for a digitally programmable amplifier in microprocessor based systems, but is also useful for discrete applications, eliminating the need to find 0.5% resistors in the ratio of 100 to 1 which track each other over temperature.

### Features

- TTL and CMOS compatible logic levels
- Microprocessor compatible
- Gain error 0.5% max
- Binary or scope knob gains
- Wide supply range +5V to  $\pm 18V$
- Packaged in 16-pin DIP

### Block Diagram and Typical Application (LF13006)



**LM1893 CARRIER-CURRENT TRANSCEIVER<sup>2</sup>**

BI-LINE<sup>1</sup>

**General Description**

Carrier-current systems use the power mains to transfer information between remote locations. This bipolar carrier-current chip performs as a power line interface for half-duplex (bi-directional) communication of serial bit streams of virtually any coding. In transmission, a sinusoidal carrier is FSK modulated and impressed on most any power line via a rugged on-chip driver. In reception, a PLL-based demodulator and impulse noise filter combine to give maximum range. A complete system may consist of the LM1893, a COPS<sup>1</sup> controller, and discrete components.

**Features**

- o Noise resistant FSK modulation
- o User-selected impulse noise filtering
- o Up to 4.8 Kbaud data transmission rate
- o Strings of 0's or 1's in data allowed
- o Sinusoidal line drive for low RFI
- o Output power easily boosted 10-fold
- o 50 to 300 KHz carrier frequency choice
- o TTL and MOS compatible digital levels
- o Regulated voltage to power logic
- o Drives all conventional power lines

**Applications**

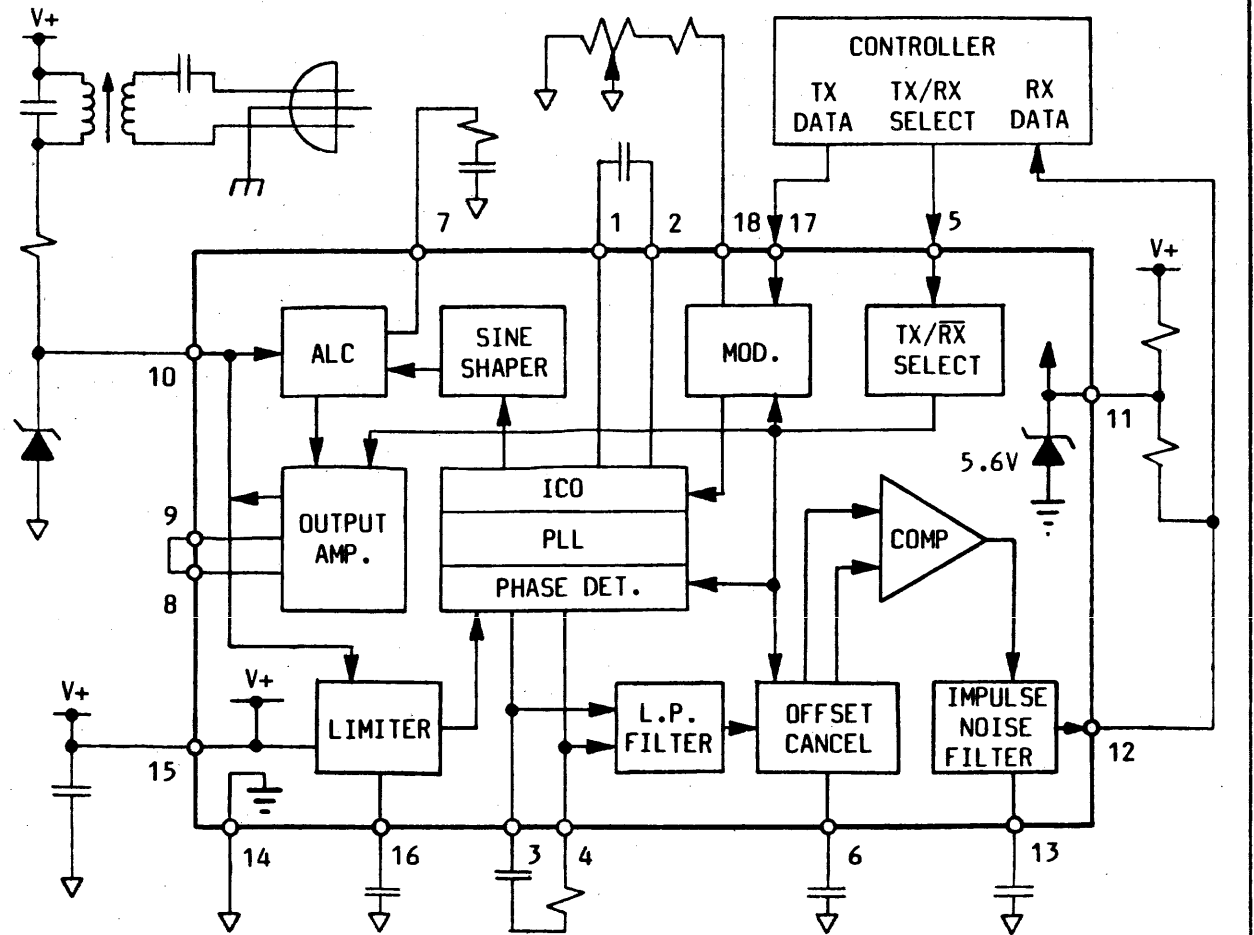
- o Energy management systems
- o Home convenience control
- o Inter-office communication
- o Appliance control
- o Fire alarm systems
- o Security systems
- o Telemetry
- o Computer terminal interface

<sup>1</sup> BI-LINE and COPS are trademarks of National Semiconductor Corp.

<sup>2</sup> Carrier-current transceivers are also called power line carrier (PLC) transceivers.

**TYPICAL APPLICATION**

FIGURE 1. Block diagram of carrier - current chip with a complement of discrete components making a complete transceiver. Use caution with this circuit - dangerous line voltage are present.



# LM2935 Low Dropout Dual Regulator

## General Description

The LM2935 positive voltage regulator features a low quiescent current of 3 mA or less when supplying 10 mA loads from the standby regulator output. This unique characteristic and the extremely low input-output differential required for proper regulation (0.55V for output currents of 10 mA) make the LM2935 the ideal regulator for power systems that include standby memory. Applications include processor power supplies demanding as much as 750 mA of output current.

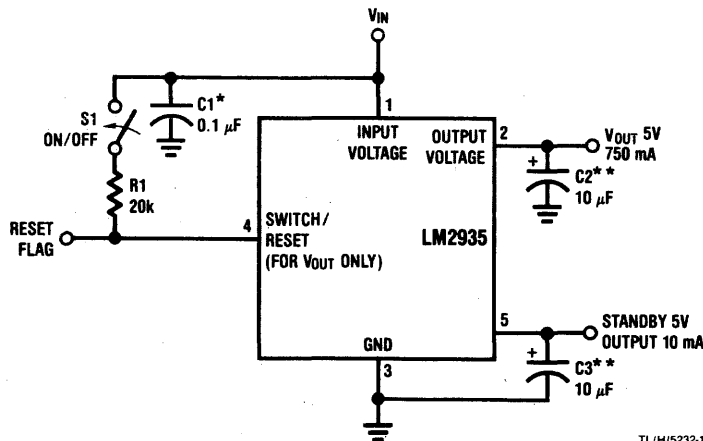
Designed primarily for automotive applications, the LM2935 and all regulated circuitry are protected from reverse battery installations or 2 battery jumps. During line transients, such as a load dump (60V) when the input voltage to the regulator can momentarily exceed the specified maximum operating voltage, the 0.75A regulator will automatically shut down to protect both internal circuits and the load while the standby regulator will continue to power any standby load. The LM2935 cannot be harmed by temporary mirror-image insertion. Familiar regulator features such as short circuit and thermal overload protection are also provided.

Fixed outputs of 5V are available in the plastic TO-220 power package.

## Features

- Two regulated outputs
- Output current in excess of 750 mA
- Low quiescent current standby regulator
- Input-output differential less than 0.6V at 0.5A
- Reverse battery protection
- 60V load dump protection
- - 50V reverse transient protection
- Short circuit protection
- Internal thermal overload protection
- Available in plastic TO-220
- ON/OFF switch for high current output
- Reset error flag
- 100% electrical burn-in

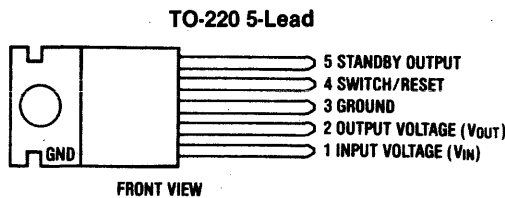
## Typical Application Circuit



\* Required if regulator is located far from power supply filter.  
\*\* Required for stability. May be increased without bound. Capacitor must be rated to operate at the minimum temperature expected for the regulator system.

FIGURE 1. Test and Application Circuit

## Connection Diagram



TL/H/5232:2

LINEAR NATIONAL SEMICONDUCTOR

# MF4 4th Order Switched Capacitor Butterworth Lowpass Filter

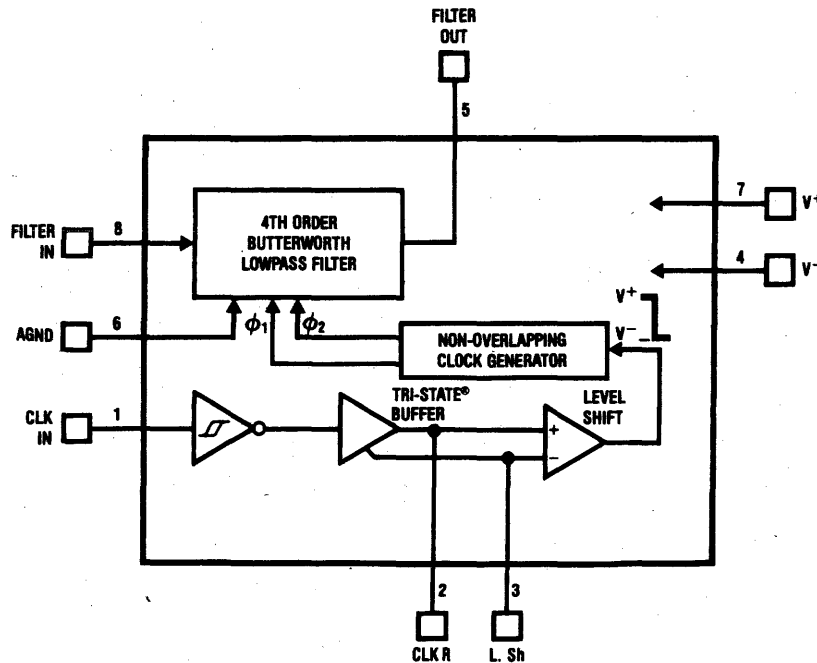
## General Description

The MF4 is a versatile, easy to use, precision 4th order Butterworth lowpass active filter. Switched capacitor techniques eliminate external component requirements and allow a clock tunable cutoff frequency. The ratio of the clock frequency to the lowpass cutoff frequency is internally set to 50 to 1 (MF4-50) or 100 to 1 (MF4-100). A Schmitt trigger clock input stage allows two clocking options, either self-clocking (via an external resistor and capacitor) for stand-alone applications, or for tighter cutoff frequency control, a TTL or CMOS logic compatible clock can be directly applied. The maximally flat passband frequency response together with a DC gain of 1 V/V allows cascading MF4 sections for higher order filtering.

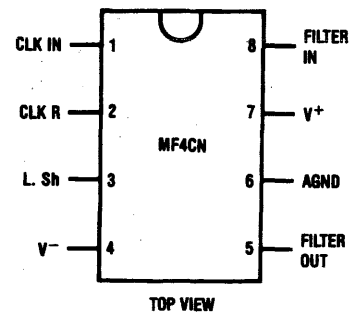
## Features

- Low Cost
- Easy to use
- No external components
- 8-pin mini-DIP
- Cutoff frequency accuracy of  $\pm 0.3\%$
- Cutoff frequency range of 0.1 Hz to 20 kHz
- 5V to 14V operation
- Cutoff frequency set by external or internal clock

## Block and Connection Diagrams



Dual-in-Line Package



TL/H/5064-2

Order Number MF4CN  
 See NS Package N08E

TL/H/5064-1

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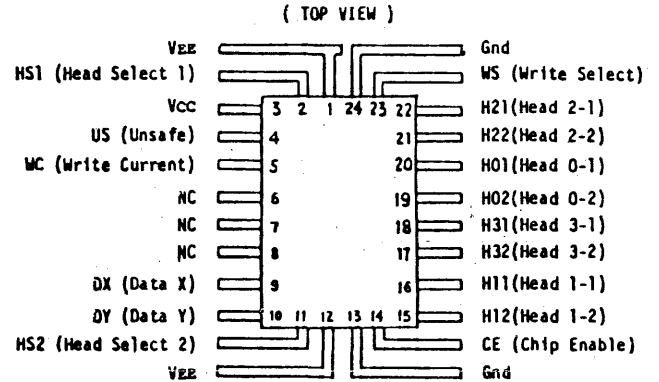
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**Read/Write IC For Disk Drive System**

The  $\mu$ PC751B/752B is a monolithic bipolar integrated circuit writing and reading data in a high performance magnetic disk storage system. The circuit is intended for use on a data head support arm. Each circuit can be used with up to four independent heads. Each I.C. contains four write drivers, four read preamplifiers, an output line driver, head select logic, and some fault monitoring circuitry, and has three modes of operation: Read, Write and Idle. The  $\mu$ PC752B is similar to the  $\mu$ PC751B in function, but provides a lower differential gain in the read mode for use with higher inductance and higher output heads. All signal levels and currents are fully compatible with IBM Model 3350 head/arm assembly.

**Pin Configuration**



**Absolute Maximum Ratings**

Parameter	Min.	Max.	Units
Positive Supply Voltage, VCC	-0.6	7.0	V
Negative Supply Voltage, VEE	-5.5	0.6	V
Input Voltage	VEE-0.3	VCC+0.3	V
Storage Temperature Range	-65	150	°C
Operating Moving-air Temperature Range	0	70	°C
Operating Junction Temperature Range		150	°C

**Recommended Operating Limits**

Parameter	Mode	Min.	Max.	Units
VCC	Read/Write/Idle	5.7	6.3	V
VEE	Read/Write/Idle	-4.2	-3.8	V
Unsafe Voltage	Read/Write/Idle	4	VCC+0.3	V
Head Center Tap Voltage	Write	3.2	3.8	V
Write Current (Iwc)	Write		45	mA

**AC Test Conditions**  
(Unless otherwise specified)

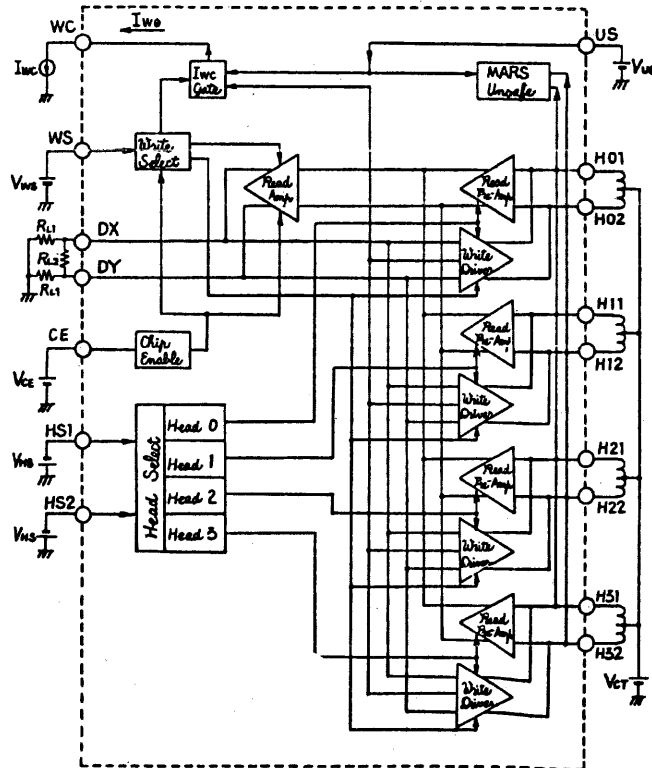
Parameter	Mode		Units
VCC	Read/Write/Idle	6.00±0.01	V
VBE	Read/Write/Idle	-4.00±0.01	V
VWS	Read	0.00±0.01	V
VWS	Write	3.50±0.01	V
VCE	Read/Write	0.00 <sup>0.01</sup> <sub>0.00</sub>	V
Lh	Read/Write/Idle	Max. 0.1	μH
Rh	Read/Write/Idle	Max. 0.01	Ω
VCT	Read	0.00±0.01	V
VCT	Write	3.50±0.01	V
VUS	Read/Write/Idle	6.00±0.01	V
RL1	Read/Write/Idle	200 ± 1	Ω
RL2	Read/Write/Idle	101 ± <sup>1</sup> / <sub>0</sub>	Ω
IWC	Read	0	mA
IWC	Write	40 ± 1	mA
Ambient Temperature	Read/Write/Idle	25 ± 2	°C

**Mode Select**

The Circuit has three modes of operation : Read, Write, and Idle. The state of the Chip Enable and Write Select determines the mode of operation as follows :

Characteristics	Conditions	Min.	Max.	Units
<b>(Chip Enable)</b>				
VHCE	Idle	5.5	6.3	V
VLCE	Read/Write	0.0	0.7	V
IHCE	Idle, VCE=VCC - 0.5V	-110	-50	μA
ILCE	Read/Write, VCE=0.0V	-1.3	-0.7	mA
<b>(Write Select)</b>				
VHWS	Write/Idle	3.2	3.8	V
VLWS	Read/Idle	-0.1	0.1	V
IHWS	Write/Idle (transition unsafe current off)	0.35	2.7	mA
	Write (transition unsafe current on)	0.7	3.5	mA
ILWS	Read/Idle		0.1	mA
Switching Delay	All Modes		0.5	μS

**Connection Diagram**



**Head Selection Table**

Head Selected	VHS1	VHS2
0	VHHS	VHHS
1	VLHS	VHHS
2	VHHS	VLHS
3	VLHS	VLHS

**Supply Current:**

Characteristics	Conditions	Min.	Max.	Units
Icc(R/W)	Read/Write	12	20	mA
Icc(ID)	Idle		150	μA
IEE(R/W)	Read/Write		70	mA
IIE(ID)	Idle		40	mA

**Total Head Input Current:**

Sum of all head input currents with Iwc=0

Characteristics	Conditions	Min.	Max.	Units
IHC(W)	Write, Vcr=3.5V		3.0	mA
IHC(R)	Read, Vcr=0.0V		0.16	mA
IHC(ID)	Idle, Vcr=3.5 or 0.0V		0.10	mA

**Head Select:**

One of four heads may be selected by controlling the state of two head select inputs. The head is selected as specified below :

Characteristics	Conditions	Min.	Max.	Units
VHHS	Temp=0°C~70°C	-0.97	-0.61	V
VLHS	Temp=0°C~70°C	-2.38	-1.52	V
IHHS			240	μA
ILHS			60	μA
Switching Delay			50	ns

**Read Mode: (μPC751B)**

In the Read mode the circuit functions as a low noise differential amplifier. The state of the head select inputs determines which amplifier is active.

Characteristics	Conditions	Min.	Max.	Units	
Input Current	Vin=0v		20	μA	
Differential Input Capacity	Vin=0v		18.8	pF	
Differential Input Resistance (Damping Resistor)	Vin=0v	25°C 0°C 70°C	585 565 585	915 915 984	Ω Ω Ω
Output Offset Voltage	Vin=0v		100	mV	
Common Mode Output Voltage	Vin=0v	-0.75	-0.50	V	
Differential Gain	Vin=1mVp-p, 0VDC f=300KHz	25°C 0°C 70°C	28 30 27	43 45 40	V/V V/V V/V
Gain Linearity (DC)	$\frac{Vo(Vin=5mV) - Vo(Vin=0mV)}{Vo(Vin=0.5mV) - Vo(Vin=0mV)} / \frac{5.0mV}{0.5mV}$		0.9	1.1	
Band Width	Zin=0Ω Vin=1mVp-p Gain @ fbw = -3dB Gain @ 300KHz		30		MHz
Input Noise	Vin=0v Zin=0Ω 15MHz Power Bandwidth		6.6		μVRMS
Common Mode Input to Differential Output Rejection Ratio	Vin=5mVp-p, 0VDC, f≤5MHz		45		dB
Power Supply Rejection Ratio (referred to the ΔVcc or ΔVEE=0.1Vp-p input)	Vin=0v, f≤5MHz		45		dB
Channel Separation	Vin=1mVp-p, 0VDC, f=300KHz, 3Channels driven		40		dB
Unsafe Current	Write Current = 0mA		0.1		mA
	Write Current = 45mA		40	45	mA

**Read Mode : (μPC752B)**

The same as μPC751B except for the following characteristics.

Characteristics	Conditions	Min.	Max.	Units	
Output Offset Voltage	Vin=0v		50	mV	
Differential Gain	Vin=1mVp-p, OVDC f=300KHz	25°C	8.5	12.5	V/V
		0°C	8.5	13.25	V/V
		70°C	7.5	12.5	V/V
Gain Linearity (DC)	$\frac{Vo(Vin=15mV) - Vo(Vin=0mV)}{Vo(Vin=0.5mV) - Vo(Vin=0mV)} / \frac{15.0mV}{0.5mV}$	0.9	1.1		
Input Noise	Vin=0v Zin=0Ω 15MHz Power Bandwidth		25	μVRMS	

**Write Mode:**

In the Write mode the circuit functions as a current gate. Externally supplied Write current is gated by the state of the head select and data inputs to one side of one head.

Characteristics	Conditions	Min.	Max.	Units
Differential Input Voltage		0.225		V
Single Ended Input Voltage		-2.0	-0.45	V
DX DY Input Current		-2.0	+2.0	mA
Current Gain	Iwc=45mA	0.95	1.0	
Write Current Voltage	Iwc=45mA	-3.7	-3.0	V
Head Current Transition Time	Iwc=45mA, Lh=0 f=5MHz		15	nS
Head Current Switching Delay Time	Iwc=45mA, Lh=0 f=5MHz		15	nS
Head Current Switching Hysteresis	Iwc=45mA, Lh=0 Data rise & fall time ≤ 1nS		2	nS
Unsafe Current	Iwc=30mA, f=2MHz, Lh=9μh Iwc=30mA, f=2MHz, Lh=0μh Iwc=45mA, Rh= one side of head only		0.1	mA
			20	mA
			20	mA
Unsafe Switching Delay Time	Iwc=30mA, f=2MHz, Lh=9μh Iwc=30mA, f=2MHz, Lh=0μh Iwc=45mA, f=2MHz, Rh=		1.0	μS
			4.0	μS
			1.0	4.0
Unselected Head Current	Iwc=45mA, f=2MHz Lh=9.5μh		1.5	mAo-p
Differential Head Voltage	Iwc=45mA, Lh=10μh		9.0	Vo-p

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**General Description**

The  $\mu$ PC754D is a two-stage wideband differential amplifier. It is intended for use as a pre-amplifier for the magnetic servo head of a magnetic disc memory system.

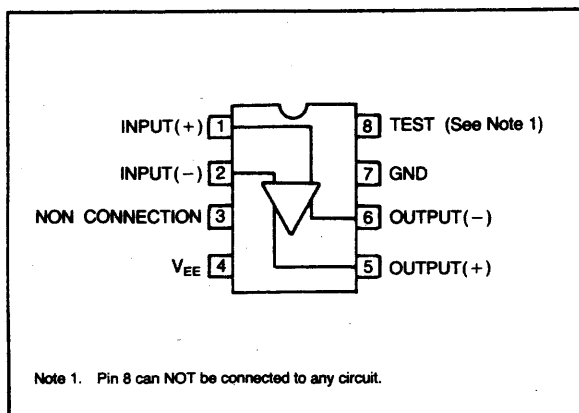
**FEATURES**

- Very narrow gain range
- 30 MHz bandwidth

**ABSOLUTE MAXIMUM RATINGS**

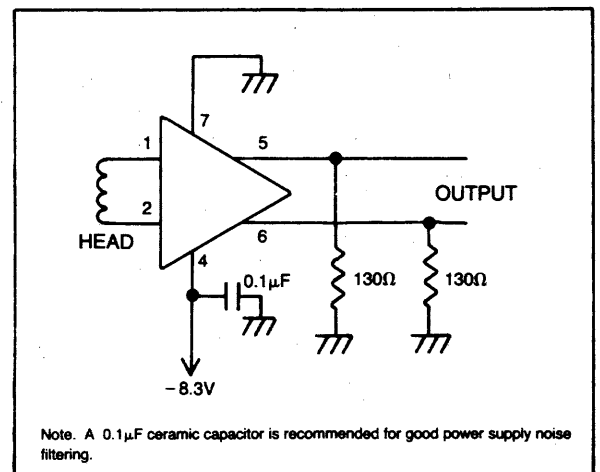
Power Supply Voltage ( $V_{EE}$ )	-12 V
Differential Input Voltage	$\pm 1$ V
Operating Temperature Range	0 to +70°C
Storage Temperature Range	-65 to +150°C

**CONNECTION DIAGRAM (Top View)**



**CONNECTION DIAGRAM**

Recommended Load Condition



**RECOMMENDED OPERATING CONDITIONS**

PARAMETER	SYMBOL	MIN	TYP	MAX	UNITS
Supply Voltage	$V_{EE}$	-7.45	-8.3	-9.15	V
Input Signal	$V_i$		2		mVpp
Ambient Temperature	$T_a$	0		70	°C

**ELECTRICAL CHARACTERISTICS ( $T_a = 25^\circ\text{C}$ ,  $V_{EE} = -8.3\text{V} \pm 10\%$ )**

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Differential Gain	$R_L = 130\Omega$	77	93	110	
Bandwidth	$V_i = 2\text{mVpp}$	10	30		MHz
Input Resistance		800	1000	1250	$\Omega$
Input Capacitance			3		pF
Output Dynamic Range (Differential)	$R_L = 130\Omega$ , $V_{EE} = -8.3\text{V}$	350			mVpp
Power Supply Current	$V_{EE} = -9.15\text{V}$		26	35	mA
Output Offset	$R_s = 0$ , $R_L = 130\Omega$			$\pm 600$	mV
Equivalent Input Noise	$R_s = 0$ , $R_L = 130\Omega$ , $\text{BW} = 4\text{MHz}$		8	14	$\mu\text{V}$
PSPR, Input Referred	$R_s = 0$ , $f \leq 5\text{MHz}$	50	65		dB
Gain Sensitivity (Supply)	$\Delta V_{EE} = \pm 10\%$ , $R_L = 130\Omega$		$\pm 1.3$		%
Gain Sensitivity (Temp)	$T_a = 25^\circ\text{C}$ to $70^\circ\text{C}$ , $R_L = 130\Omega$		-0.2		%/°C
CMRR, Input Referred	$f \leq 5\text{MHz}$	55	.70		dB

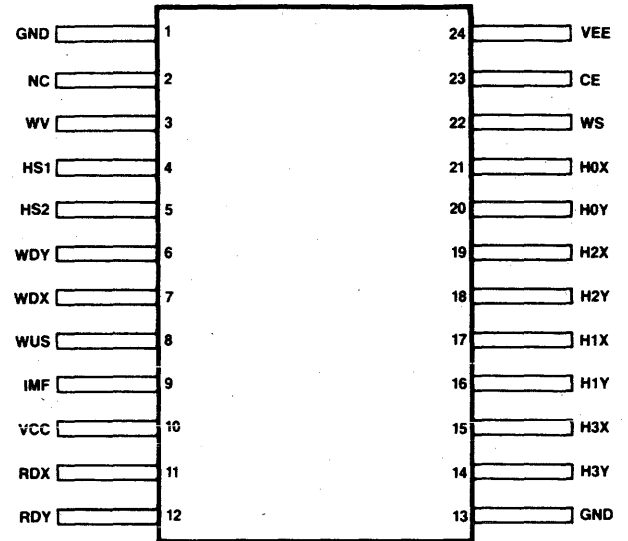
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**Description**

The circuit is an integrated Read/Write chip intended for use with non-center tapped thin film heads in disk drive systems. Each chip controls four heads, and has three modes of operation: read, write and idle.

The chip has four channels of read amplifiers and write drivers. It also has an internal write current source. The write current magnitude is determined by laser trimmed internal resistors. The chip contains fault detection circuits.

**PIN CONFIGURATION (Top View)**



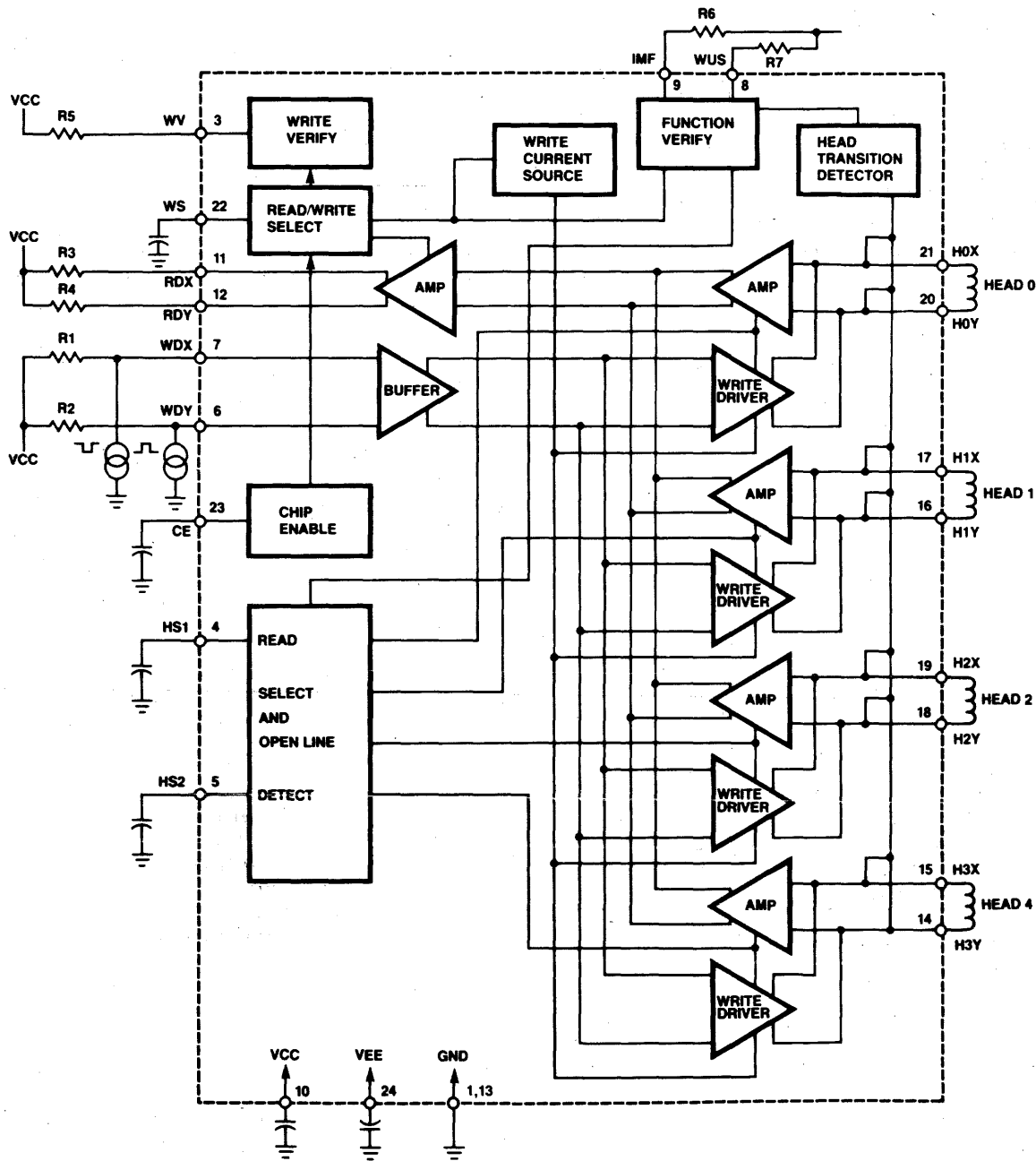
**PIN NAMES AND DEFINITIONS**

- CE = Chip Enable
- WS = Write Select
- HS1 = Head Select 1
- HS2 = Head Select 2
- WDX = Write Data X
- WDY = Write Data Y
- RDX = Read Data X
- RDY = Read Data Y
- WV = Write Verify (write Select OK)
- IMF = Input Multi Fault
- WUS = Write Unsafe (Write Operation OK)

**RECOMMENDED OPERATING LIMITS**

PARAMETER	MODE	MIN	MAX	UNIT
VCC	Read/Write/Idle	4.75	5.25	V
VEE	Read/Write/Idle	-5.46	-4.75	V
Write Mode Duty			50	%

**BLOCK DIAGRAM**



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**PIN NAMES AND DEFINITIONS**

- CE = Chip Enable
- WS = Write Select
- HS1 = Head Select 1
- HS2 = Head Select 2
- WDX = Write Data X
- WDY = Write Data Y
- RDX = Read Data X
- RDY = Read Data Y
- WV = Write Verify (write Select OK)
- IMF = Input Multi Fault
- WUS = Write Unsafe (Write Operation OK)

**RESISTOR VOLTAGE**

- R1 = 100Ω ± 1%
- R2 = 100Ω ± 1%
- R3 = 100Ω ± 1%
- R4 = 100Ω ± 1%
- R5 = 1k ± 1%
- R6 = 330Ω ± 1%
- R7 = 1k ± 1%



**ELECTRICAL SPECIFICATIONS**

Unless otherwise specified,  $4.75\text{ V} \leq V_{CC} \leq 5.25\text{ V}$ ,  
 $-5.46\text{ V} \leq V_{EE} \leq -4.75\text{ V}$ ,  $25^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

**POWER SUPPLY CURRENT**

CHARACTERISTICS		CONDITION	TYP	UNIT
I <sub>CC</sub>	Idle	V <sub>CE</sub> = 2V		mA
	Read	V <sub>CE</sub> = 0V, V <sub>WS</sub> = 2V	47	mA
	Write	V <sub>CE</sub> = 0V, V <sub>WS</sub> = 0V	37	mA
I <sub>EE</sub>	Idle	V <sub>CE</sub> = 2V		mA
	Read	V <sub>CE</sub> = 0V, V <sub>WS</sub> = 2V	50	mA
	Write	V <sub>CE</sub> = 0V, V <sub>WS</sub> = 0V	120	mA

**HEAD SELECT**

CHARACTERISTICS	CONDITION	MIN	MAX	UNIT
V <sub>HS1H</sub> , V <sub>HS2H</sub>		2.0		V
V <sub>HS1L</sub> , V <sub>HS2L</sub>		-0.3	0.8	V
I <sub>HS1H</sub> , I <sub>HS2H</sub>			50	μA
I <sub>HS1L</sub> , I <sub>HS2L</sub>		-1.6		mA
Switching Delay	All Mode		0.4	μS

**HEAD SELECTION TABLE**

HEAD SELECTED	V <sub>HS1</sub>	V <sub>HS2</sub>
0	L	L
1	H	L
2	L	H
3	H	H

L = LOW level H = HIGH level

**MODE SELECT**

CHARACTERISTICS	CONDITION	MIN	MAX	UNIT
Chip Enable				
V <sub>CEH</sub>	Idle	2.0		V
V <sub>CEL</sub>	Read or Write	-0.3	0.8	V
I <sub>CEH</sub>	Idle V <sub>CE</sub> = 5.0V		50	μA
I <sub>CEL</sub>	Read or Write, V <sub>CE</sub> = 0V	-4.0		mA
Switching Delay	All Mode		0.5	μS
Write Select				
V <sub>WSH</sub>	Read or Idle	2.0		V
V <sub>WSL</sub>	Write or Idle	-0.3	0.8	V
I <sub>WSH</sub>	Read or Idle		50	μA
I <sub>WSL</sub>	Write or Idle	-1.6		mA
Switching Delay	All Mode		0.5	μS

**WRITE MODE**

CHARACTERISTICS	CONDITION	MIN	MAX	UNIT
Differential Input Voltage		0.2		V
Single-ended Input Voltage		-1.87	0.1	V
Input Current/Side		-20	120	$\mu$ A
Input Capacitance/Side			10	pF
Write Current		67.9	72.1	mA o-p
Write Current Tolerance		-3	+3	%
Write Current Transition Time	$I_{WC} = 70$ mA o-p, $R_H = 8 \Omega$ $L_H = 0.1 \mu$ H		17	nS
Write Current Switching Delay Time	$I_{WC} = 70$ mA o-p, $L_H = 0 \mu$ H		15	nS
Write Current Switching Delay Time Asymmetry	$I_{WC} = 70$ mA o-p, $L_H = 0 \mu$ H		1	nS
Unselected Head Current	$I_{WC} = 100$ mA o-p, $f = 5$ MHz $R_H = 8 \Omega$ , $L_H = 0.1 \mu$ H		2	mA o-p
Output Resistance		840	1260	$\Omega$
Output Capacitance			30	pF
Differential Head Voltage Swing			6	V o-p

**READ MODE**

CHARACTERISTICS	CONDITION	MIN	MAX	UNIT
Input Bias Current	$V_{IN} = 0$ V		36	$\mu$ A
Input Offset Current	$V_{IN} = 0$ V		20	$\mu$ A
Differential Input Resistance	$V_{IN} = 0$ V	520	1170	$\Omega$
Differential Input Capacitance	$V_{IN} = 0$ V		40	pF
Differential Input Noise Voltage	$V_{IN} = 0$ V, $Z_{IN} = 0 \Omega$ , $B_W = 15$ MHz		6.6	$\mu$ Vrms
Differential Voltage Gain	$V_{IN} = 1$ mV p-p, $f = 1$ MHz			
	$T_A = 25^\circ$ C	100	150	V/V
	$T_A = 0^\circ$ C			V/V
	$T_A = 70^\circ$ C			V/V
Input Dynamic Range	Gain Is 90% of the Gain With 0.5 mV p-p Input Signal $f = 1$ MHz	6		mV p-p
Voltage Band Width	$V_{IN} = 1$ mV p-p	$\pm 1$ dB	25	MHz
		-3 dB	45	MHz
CMRR	$V_{IN} = 5$ mV p-p	$f = 10$ MHz	54	dB
		( $f = 20$ MHz)	48	dB
PSRR	$V_{IN} = 0$ mV	$f = 10$ MHz	58	dB
		( $f = 20$ MHz)	50	dB
Channel Separation	$V_{IN} = 0$ mV 3 Channel Driven	$f = 10$ MHz	40	dB
Output CM Voltage	$R_L = 100 \Omega$ to $V_{CC}$ $T_j = 75^\circ$ C	$V_{CC} - 0.75$	$V_{CC} - 0.50$	V
Output Offset Voltage	$R_L = 100 \Omega$ to $V_{CC}$	-360	360	mV
Output Leak Current	$V_{CE} = 2$ V (Idle)		10	$\mu$ A
Output Resistance	Single-ended	10		k $\Omega$
Output Capacitance	Single-ended		10	pF

**FAULT DETECTION**

CHARACTERISTICS	CONDITION	MIN	MAX	UNIT
I <sub>IMFH</sub>	V <sub>CE</sub> = 0 V, Read or Write	2.85	3.15	mA
I <sub>IMFL</sub>	V <sub>CE</sub> = 2 V, Idle		20	μA
IMF Switching Delay Time	Chip Enable ON/OFF		0.5	μS
V <sub>WVL</sub> Write Select OK	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 0 V I <sub>WV</sub> = 16 mA	0	0.4	V
I <sub>WVH</sub> Write Select OK	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 2 V	0	250	μA
	V <sub>CE</sub> = 2 V	0	250	μA
WV Switching Delay Time	Write Mode ON/OFF		0.5	μS
I <sub>WUSH</sub> Write Operation OK	V <sub>CE</sub> = 2 V	0	250	μA
	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 2 V	0	250	μA
	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 0 V L <sub>H</sub> = 0 μH, R <sub>H</sub> = 0 Ω f (Write Date) = 5 MHz	0	250	μA
	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 0 V R <sub>H</sub> = ∞ f (Write Date) = 5 MHz	0	250	μA
	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 0 V L <sub>H</sub> = 0.1 μH, R <sub>H</sub> = 8 Ω f (Write Date) = 200 kHz	0	250	μA
	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 0 V L <sub>H</sub> = 0.1 μH, R <sub>H</sub> = 8 Ω f (Write Date) = 5 MHz Head Select Line → Open	0	250	μA
V <sub>WUSL</sub> Write Operation OK	V <sub>CE</sub> = 0 V, V <sub>WS</sub> = 0 V L <sub>H</sub> = 0.05 μH, R <sub>H</sub> = 5 Ω f (Write Date) = 1 to 10 MHz	0	0.4	V
WUS Switching Delay Time	Unsafe → Safe		1	μS
	Safe → Unsafe	(1)	3.6	μS

**Description**

The  $\mu$ PC1651G is a silicon monolithic integrated circuit especially designed as wide band amplifier covered from HF band to UHF band.

**Features**

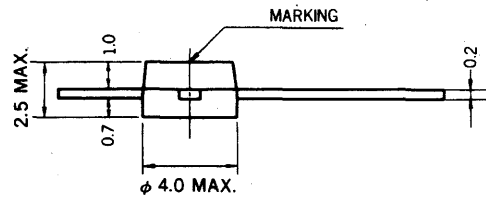
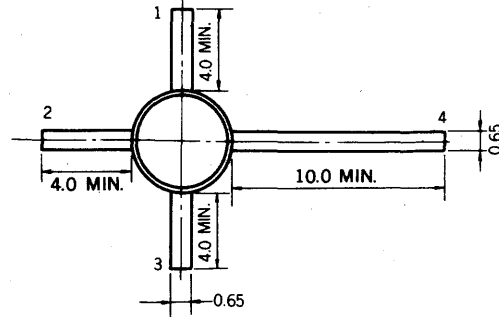
- Excellent frequency response; TYP. 1200 MHz @3 dB down
- High power gain; TYP. 19 dB @f = 500 MHz
- Low voltage operation; 5V
- Small package

**ABSOLUTE MAXIMUM RATINGS ( $T_a = 25^\circ\text{C}$ )**

Supply Voltage	$V_{CC}$	6	V
Total Power Dissipation	$P_T$	250	mW
Operating Temperature	$T_{opt}$	-20 to +75	$^\circ\text{C}$
Storage Temperature	$T_{stg}$	-40 to +125	$^\circ\text{C}$

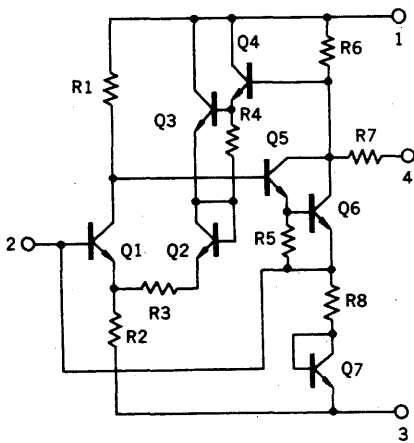
**PACKAGE DIMENSIONS**

in millimeters



1.  $V_{CC}$
2. INPUT
3. GND
4. OUTPUT

**EQUIVALENT CIRCUIT**

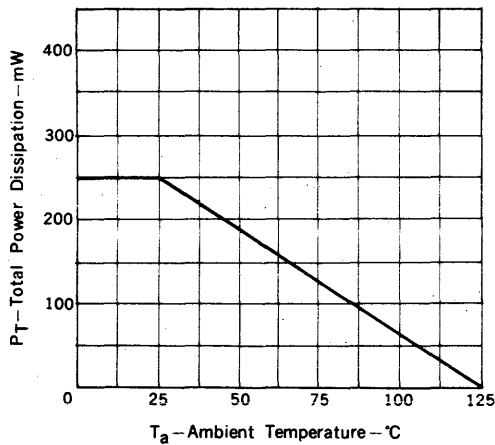


**ELECTRICAL CHARACTERISTICS ( $T_a = 25^\circ\text{C}$ )**

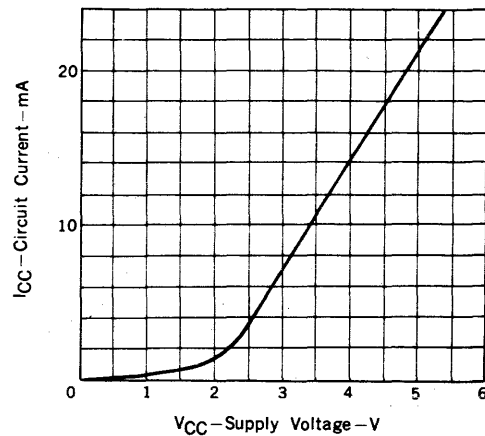
CHARACTERISTIC	SYMBOL	MIN.	TYP.	MAX.	UNIT	TEST CONDITIONS
Circuit Current	$I_{CC}$	15	20	25	mA	$V_{CC} = 5\text{ V}$
Gain	$G_V$	16	19		dB	$V_{CC} = 5\text{ V}, f = 500\text{ MHz}$
Noise Figure	NF		5.5	6.5	dB	$V_{CC} = 5\text{ V}, f = 500\text{ MHz}$
Band Width	BW	1 000	1 200		MHz	$V_{CC} = 5\text{ V}, 3\text{ dB down}$
Isolation	$I_{so}$	20	24		dB	$V_{CC} = 5\text{ V}, f = 500\text{ MHz}$
Input Return Loss	$ S_{11} $		15		dB	$V_{CC} = 5\text{ V}, f = 500\text{ MHz}$
Output Return Loss	$ S_{22} $		10		dB	$V_{CC} = 5\text{ V}, f = 500\text{ MHz}$
Maximum Output Level	VOM		112		dB $\mu$ V	$V_{CC} = 5\text{ V}, f = 500\text{ MHz}$

TYPICAL CHARACTERISTICS ( $T_a = 25^\circ\text{C}$ )

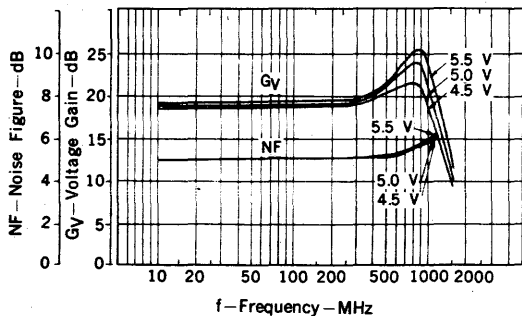
TOTAL POWER DISSIPATION vs. AMBIENT TEMPERATURE



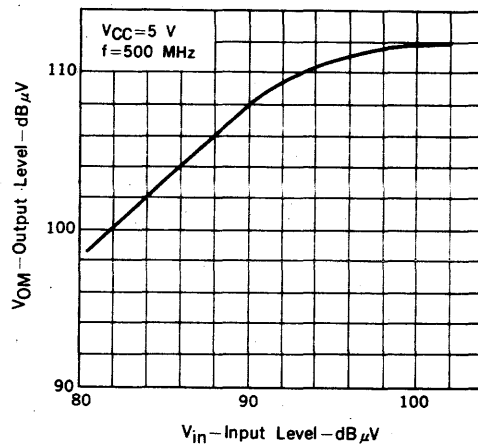
CIRCUIT CURRENT vs. SUPPLY VOLTAGE



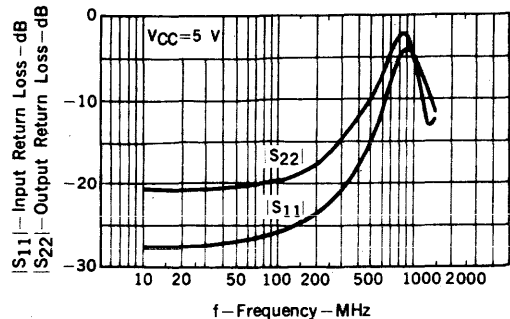
NOISE FIGURE AND VOLTAGE GAIN vs. FREQUENCY



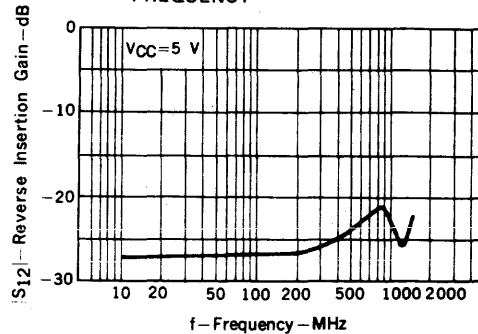
OUTPUT LEVEL vs. INPUT LEVEL



INPUT AND OUTPUT RETURN LOSS vs. FREQUENCY



REVERSE INSERTION GAIN vs. FREQUENCY



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## Linear IC (OP AMPS/Comparators/Voltage Regulators)

### Operational Amplifiers

				Package	Supply Voltage (V)	Power Consumption (mW)	Input Offset Voltage MAX (mV)	Input Offset Current MAX (nA)	Input Bias Current MAX (nA)	Output Voltage MIN (V)	Slew Rate TYP (V/μs)	Equiv. Vn (Input) TYP (μVrms)
Dual Power Supply	Low Noise Types	Dual	AN6550	9-SIP	±2 to ±12 4 to 24	15	6	200	500	±1	0.8	2.5
			AN6551		±4 to ±15 8 to 30	170	6	200	500	±10	1.0	2.5
			AN6555		±4 to ±15 8 to 30	170	6	200	500	±10	2.0	1.5
		AN6557	9-SIP(LP)	±4 to ±15 8 to 30	240	3	200	—	±10	6.0	0.9	
		AN6558 (AN6552)	8-DIP	±4 to ±15 8 to 30	170	6	200	500	±10	1.0	2.5	
		AN6553		±4 to ±15 8 to 30	170	6	200	500	±10	2.0	2.5	
		AN6556		±4 to ±15 8 to 30	170	6	200	500	±10	2.0	1.5	
		AN6558		±4 to ±15 8 to 30	240	3	200	—	±10	6.0	0.9	
	AN6558S	SO-8D	±4 to ±15 8 to 30	170	6	200	500	±10	1.0	2.5		
	AN6556S		±4 to ±15 8 to 30	170	6	200	500	±10	2.0	.5		
	AN6554	14-DIP	Quad	±2 to ±15 4 to 30	240	5	50	300	±10	1.6	2.5	
	AN6554NS	SO-14D		±2 to ±15 4 to 30	240	5	50	300	±10	1.6	2.5	
	General Purpose	Single	AN6573	7-SIP	±2 to ±15 4 to 30	85	4	100	250	±10	0.7	4.0
			AN6593	9-SIP(LP)	±1 to ±18 2 to 36	3	6	20	75	±12	—	—
AN1741 (AN6570)			8-DIP	±2 to ±15 4 to 30	85	4	100	250	±10	0.7	4.0	
AN4250				±1 to ±18 2 to 36	3	6	20	75	±12	—	—	
AN1741S			SO-8D	±2 to ±15 4 to 30	85	4	100	250	±10	0.7	4.0	
AN4250S			SO-8D	±1 to ±18 2 to 36	3	6	20	75	±12	—	—	
AN6571		9-SIP(LP)	Dual	±2 to ±15 4 to 30	170	4	100	250	±10	0.7	4.0	
AN1458 (AN6572)		8-DIP	±2 to ±15 4 to 30	170	4	100	250	±10	0.7	4.0		
AN1458S			SO-8D	±2 to ±15 4 to 30	170	4	100	250	±10	0.7	4.0	
Single Power Supply		General Purpose	Dual	AN6561	9-SIP	±1.5 to ±15 3 to 30	6	7	50	250	V <sub>CC</sub> -1.5	0.3
	AN1358 (AN6562)			8-DIP	±1.5 to ±15 3 to 30	6	7	50	250	V <sub>CC</sub> -1.5	0.3	6.0
	AN1358S			SO-8D	±1.5 to ±15 3 to 30	6	7	50	250	V <sub>CC</sub> -1.5	0.3	6.0
	AN1324 (AN6564)		Quad	±1.5 to ±15 3 to 30	10	7	50	250	V <sub>CC</sub> -1.5	0.3	6.0	
	AN1324NS			SO-14D	±1.5 to ±15 3 to 30	10	7	50	250	V <sub>CC</sub> -1.5	0.3	6.0

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## Linear IC (OP AMPS/Comparators/Voltage Regulators)

### Comparators

		Package	Supply Voltage (V)		Supply Current MAX (mA)	Input Offset Voltage MAX (mV)	Input Offset Current MAX (nA)	Input Bias Current MAX (nA)	Output Current MIN (mA)	Response Time TYP (ms)
Dual	AN6913	9-SIP	$\pm 1$ to $\pm 18$	2 to 36	1.5	5	50	250	10	1.3
	AN6915		$\pm 1$ to $\pm 18$	2 to 36	5.3	5	50	200	70	2.0
	AN1393 (AN6914)	8-DIP	$\pm 1$ to $\pm 18$	2 to 36	1.5	5	50	250	10	1.3
	AN6916		$\pm 1$ to $\pm 18$	2 to 36	5.3	5	50	200	70	2.0
	AN1393S	SO-8D	$\pm 1$ to $\pm 18$	2 to 36	1.5	5	50	250	10	1.3
	AN6916S		$\pm 1$ to $\pm 18$	2 to 36	5.8	5	50	200	70	2.0
Quad-ruple	AN1339 (AN6912)	14-DIP	$\pm 1$ to $\pm 18$	2 to 36	1.5	5	50	250	6	1.3
	AN6918		$\pm 1$ to $\pm 18$	2 to 36	1.5	5	50	250	6	1.3
			$\pm 1$ to $\pm 18$	2 to 36	10.0	5	50	200	70	2.0
	AN1339S	SO-14D	$\pm 1$ to $\pm 18$	2 to 36	1.5	5	50	250	6	1.3

### Voltage Regulators

#### Positive Output 3 Terminals (AN7800/AN78M00/AN78L00 Series)

I <sub>o</sub>	Output Voltage (V)											
	4	5	6	7	8	9	10	12	15	18	20	24
1A	—	AN7805	AN7806	AN7807	AN7808	AN7809	AN7810	AN7812	AN7815	AN7818	AN7820	AN7824
0.5A	—	AN78M05	AN78M06	AN78M07	AN78M08	AN78M09	AN78M10	AN78M12	AN78M15	AN78M18	AN78M20	AN78M24
0.1A	AN78L04	AN78L05	AN78L06	AN78L07	AN78L08	AN78L09	AN78L10	AN78L12	AN78L15	AN78L18	AN78L20	AN78L24

Package: AN7800/AN78M00 Series = TO-220, AN78L00 Series = TO-92

#### Negative Output 3 Terminals (AN7900 Series)

I <sub>o</sub>	Output Voltage (V)										
	-5	-6	-7	-8	-9	-10	-12	-15	-18	-20	-24
1A	AN7905	AN7906	AN7907	AN7908	AN7909	AN7910	AN7912	AN7915	AN7918	AN7920	AN7924

Package: TO-220

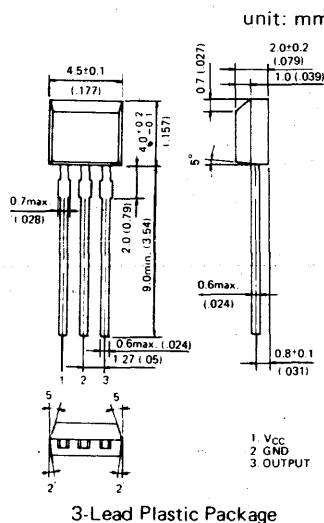
## Hall IC Series

### Specifications

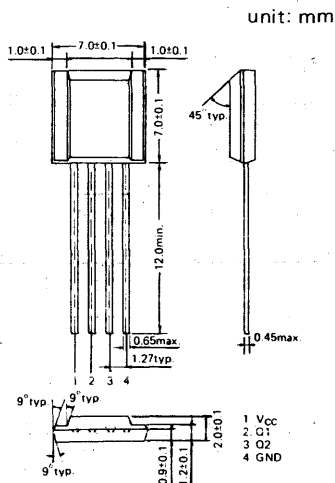
#### Selection Guide

Package	Type No.	Type	Supply Voltage (V)	Operating Ambient Temp. Range (°C)	Magnetic Flux Density (Gauss)	Output Current (mA) (Value for One Output)	Output
4-Lead	DN834	Switch Type	4.5 to 5.5	-20 to +75	100 to 750	15	DTL
	DN835	Linear Type	4.5 to 5.5	-20 to +75	-350 to 350	-15 to 4.4	Emitter Follower
	DN837	Switch Type	4.5 to 5.5	-20 to +75	100 to 750	15	Open Collector
	DN838	Switch Type	8 to 16	-40 to +100	-300 to 300	-1 to 20	DTL
	DN839	Switch Type	4.5 to 16	-20 to +75	100 to 750	20	Open Collector
4-Lead	DN6835	Linear Type	4.5 to 5.5	-20 to +75	-500 to 500	-15 to 4.4	Emitter Follower
	DN6836	Linear Type	4.5 to 5.5	-20 to +75	-500 to 500	-15 to 4.4	Emitter Follower
	DN6837	Switch Type	4.5 to 5.5	-20 to +75	100 to 750	15	Open Collector
	DN6838	Switch Type	8 to 16	-40 to +100	-300 to 300	20	DTL
	DN6839	Switch Type	4.5 to 16	-20 to +75	100 to 750	20	Open Collector

### Dimensions



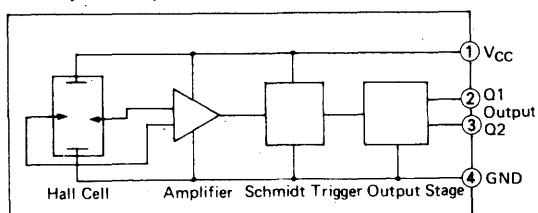
3-Lead Plastic Package



4-Lead Plastic Package

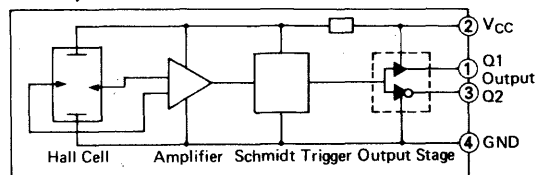
### Circuit Block Diagram

#### DN834, DN837, DN6837



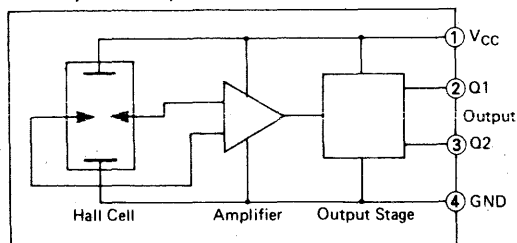
Note: DN6837 V<sub>CC</sub> = (1), Output = only (3)  
GND = (2)

#### DN838, DN6838



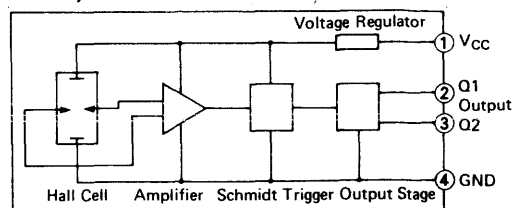
Note: DN6838 V<sub>CC</sub> = (1), Output = only (3),  
GND = (2)

#### DN835, DN6835, DN6836



Note: DN6835, DN6836 V<sub>CC</sub> = (1)  
Output = only (3) GND = (2)

#### DN839, DN6839



Note: DN6839 V<sub>CC</sub> = (1), Output = only (3),  
GND = (2)

Specifications subject to change without notice.



# BBD (Bucket Brigade Devices)

## BBD QUICK REFERENCE TABLE

Item	Symbol	MN3001	MN3002	MN3003	MN3004	MN3005	MN3007	MN3007	MN3008	MN3009	MN3010	MN3011	MN3012	Unit
Number of BBD	N	Dual-512	512	Dual-64	512	4096	128	1024	2048	256	Dual-512	3328 6-Tap	190, 3, 5 3-Tap	Stage
Circuit Construction	Clock Generator	External		Built-in	External								Built-in	
	Output Terminal	Pair										6 Different Taps	3 Different Taps	
Operating Conditions	Drain Supply Voltage	V <sub>DD</sub>	-15	-9	-15	-15	-15	-15	-15	-15	-15	-15	-15	V
	Gate Supply Voltage	V <sub>GG</sub>	-14	-8	-14	-14	-14	-14	-14	-14	-14	-14	-	V
	Back-Gate Bias Voltage	V <sub>BB</sub>	+5	Not Needed										V
	Clock Voltage "H" Level	V <sub>CPH</sub>	0	0	0	0~-1	0~-1	0~-1	0~-1	0~-1	0~-1	0~-1.3	-0~-0.4	V
	Clock Voltage "L" Level	V <sub>CPL</sub>	-15	-9	-15	-15	-15	-15	-15	-15	-15	-15	-15	V
	Input DC Bias	V <sub>Bias</sub>	-3.3~-4.9	-2.5~-6	-5~-10	-5~-10	-5~-10	-5~-10	-5~-10	-5~-10	-5~-10	-5~-10	-	-3~-12
Electrical Characteristics	Input Signal Frequency	f <sub>i</sub>	12			10	12			14	12	10	12	kHz (max.)
	Input Signal Swing	V <sub>i</sub>	1.8	0.8	1.8	1.2	1.8	1.5	1.5	1.7	1.8	1.0	1.2	V <sub>rms</sub> (max.)
	Insertion Loss	L <sub>i</sub>	8.5	3.5	1.5	0	0	0	0	0	0	0	0	dB (typ.)
	Total Harmonic Distortion	THD	0.4	0.5	0.4	1	0.2	0.5	0.5	0.3	0.4	0.4	0.4	% (typ.)
	Noise Voltage	V <sub>no</sub>	250 (typ.)	140	210	400	200	300	300	150	210	0.4	0.14	μV <sub>rms</sub> (max.)
	Signal to Noise Ratio	S/N	70	75	85	75	90	80	78	88	85	76	OUT1=90 OUT2=97 OUT3=98	dB (typ.)
	Signal Delay Time	t <sub>D</sub>	51.2	25.6	6.4	25.6	204.8	6.4	51.2	102.4	12.8	51.2	t <sub>D1</sub> =19.8/33.1 59.7/86.3 139.5/166.4	0.475~9.5 0.0125~0.25 0.0075~0.15
Package (Molded Package)		14-Pin DIP				Larger 8-Pin DIP	8-Pin DIP		Larger 8-Pin DIP	8-Pin DIP	14-Pin DIP	12-Pin DIP	14-Pin DIP	

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# SP8612B

1800MHz ÷ 4

The SP8612B is an asynchronous emitter coupled logic counter which provides ECL 10K compatible outputs and can drive 100 ohm lines. They operate from a -6.8V or +5, -1.8V supply but are otherwise similar to the SP8610 and 8611.

### FEATURES

- ECL Compatible Output
- AC Coupled Input (Internal Bias)
- Typical Operating Frequency 2GHz

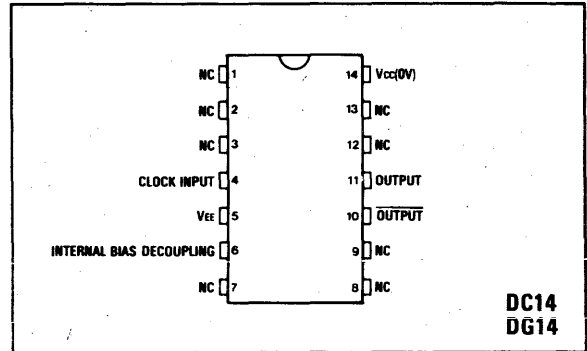


Fig.1 Pin connections - top view

### QUICK REFERENCE DATA

- Supply Voltage: -6.8V
- Power Consumption: 500mW
- Output Voltage Swing: 60mV typ.

### ABSOLUTE MAXIMUM RATINGS

- Supply Voltage: (V<sub>CC</sub>-V<sub>EE</sub>) 8V
- Output Current: 15mA
- Storage Temperature Range: -55°C to +150°C
- Max. Junction Temperature: +150°C
- Max. Input Voltage: 2.5V p-p

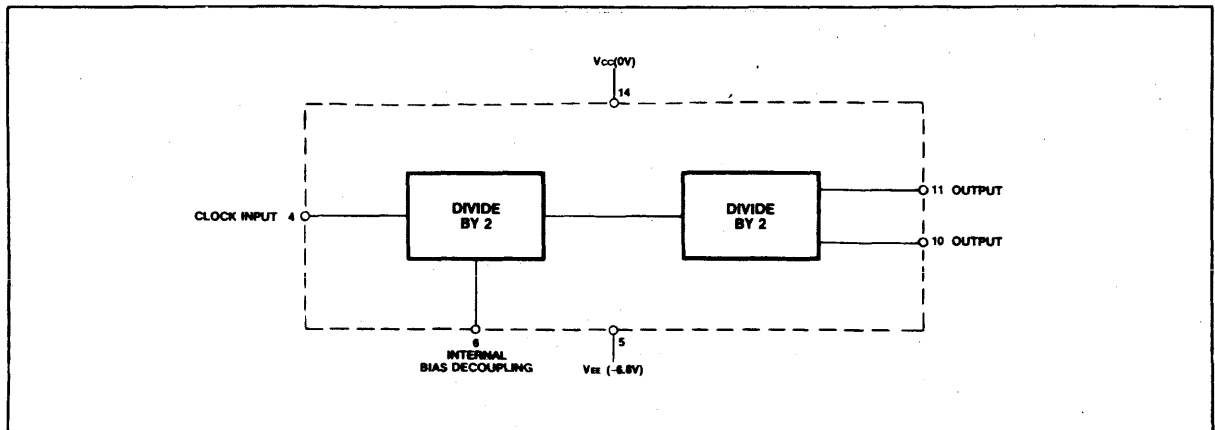


Fig.2 Functional diagram

**ELECTRICAL CHARACTERISTICS**

 Supply voltage:  $V_{CC} = 0V$ ,  $V_{EE} = -6.8V \pm 0.25V$   
 Temperature:  $T_{amb} = 0^{\circ}C$  to  $+70^{\circ}C$ 

Characteristics	Symbol	Value		Units	Conditions	Temperature
		Min.	Max.			
Maximum frequency sinewave input	$f_{max}$	1.8		GHz	Input = 800-1200mV p-p	As specified
Minimum frequency sinewave input	$f_{min}$		900	MHz	Input = 400-1200mV p-p	As specified
Power supply current	$I_{EE}$		110	mA	Outputs unloaded $V_{EE} = -7.15V$	As specified
Output low voltage	$V_{OL}$	-1.9	-1.62	V	Outputs loaded with $620\Omega$ to $V_{EE} = -6.8V$	$25^{\circ}C$
Output high voltage	$V_{OH}$	-0.93	-0.75	V	Outputs loaded with $620\Omega$ to $V_{EE} + -6.8V$	$25^{\circ}C$
Minimum output swing	$V_{OUT}$	500		mV	Outputs loaded with $620\Omega$ to $V_{EE} = -6.8V$	As specified

**NOTES**

1. Unless otherwise stated, the electrical characteristics are guaranteed over specified supply, frequency and temperature range.
2. The temperature coefficients of  $V_{OH} = +1.2mV/^{\circ}C$  and  $V_{OL} = +0.24mV/^{\circ}C$ .
3. The test configuration for dynamic testing is shown in Fig. 5.

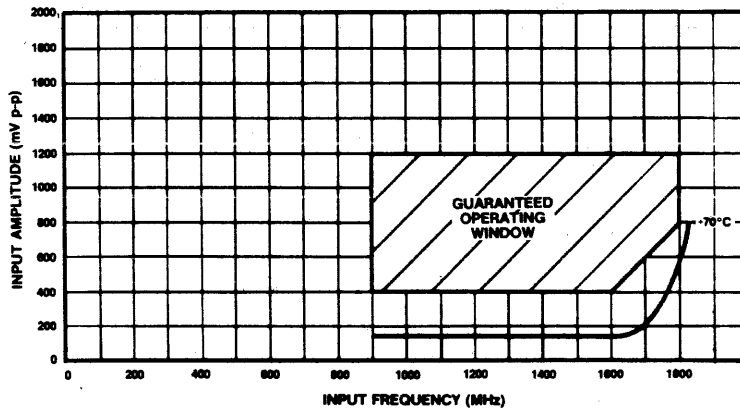


Fig.3 Typical input characteristics SP8612B

**OPERATING NOTES**

1. The clock input (pin 4) should be capacitively coupled to the signal source. The input signal path is completed by connecting a capacitor from the internal bias decoupling, pin 6 to ground.
2. If no signal is present the device will self-oscillate. If this is undesirable it may be prevented by connecting a 10k resistor from the input to  $V_{EE}$  (i.e. pin 4 to pin 7). This reduces sensitivity by approximately 100mV.
3. The input can be operated at very low frequencies but slew rate must be better than  $200V/\mu s$ .
4. The input impedance of the SP8612 is a function of

frequency. See Fig. 4.

5. The emitter follower outputs require external load resistors. These should not be less than 330 ohms, and a value of 620 ohms is recommended. Interfacing to ECL III/10K is shown in Fig. 7.

6. These devices may be used with split supply lines by means of the circuit of Fig. 6. Some improvement in the upper frequency of operation may be obtained under these conditions, but suitable circuit layout must be employed to achieve this improvement.

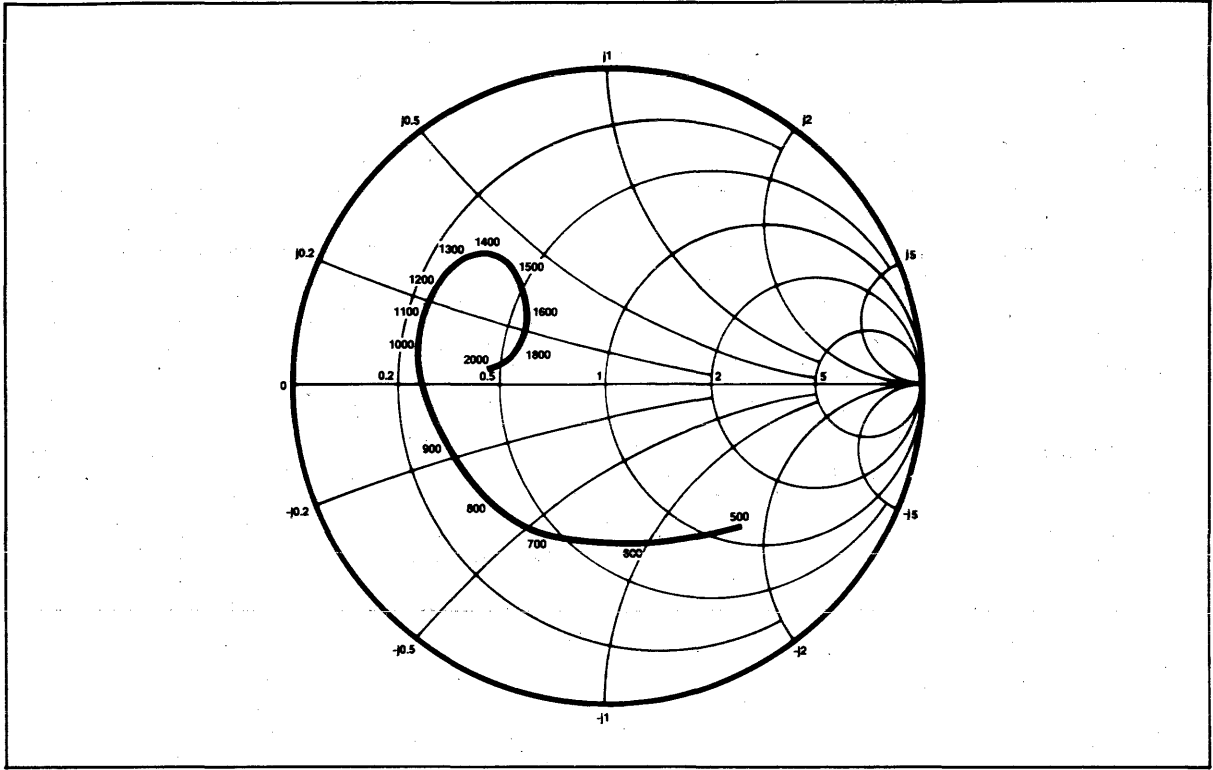


Fig.4 Typical input impedance. Test conditions: supply voltage  $\pm 6.8V$ , ambient temperature  $25^{\circ}C$ . Frequencies in MHz, impedances normalised to 50 ohms.

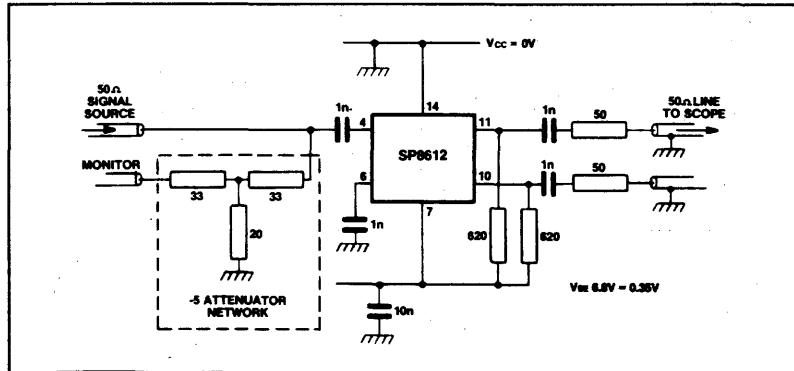


Fig.5 Toggle frequency test circuit

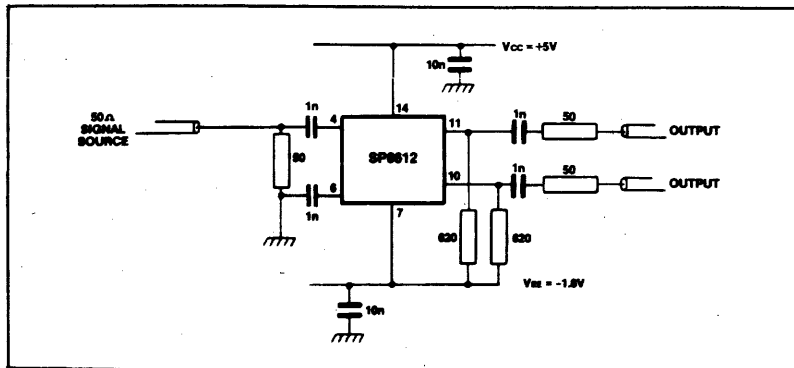


Fig.6 Operation on split supply voltages

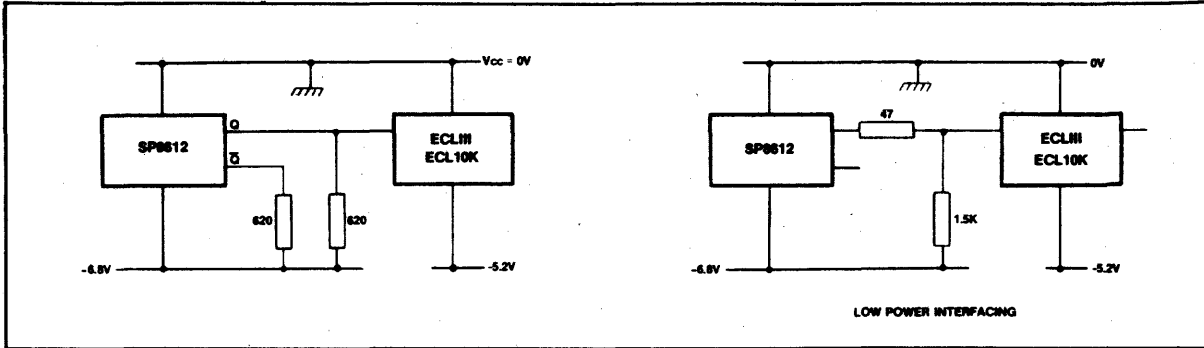


Fig.7 Interfacing SP8612 series to ECL 10K and ECL III

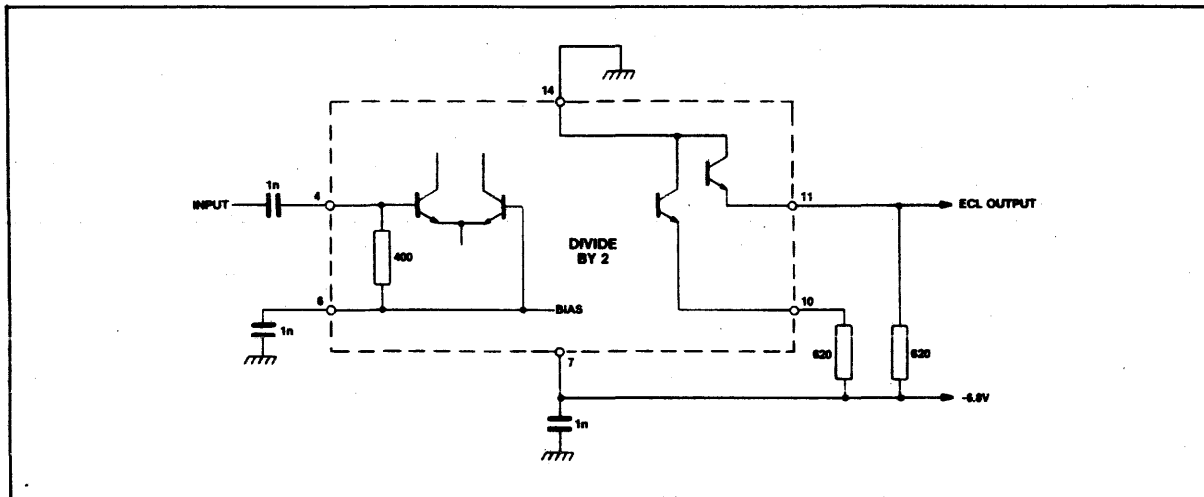


Fig.8 Typical application showing interfacing

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## OPERATIONAL AMPLIFIERS

### SELECTION GUIDE

The Operational Amplifier Selection Guide chart highlights PMI's line of operational amplifiers. The matrix indicates the most essential parametric differences for each product group.

"General Purpose" op amps are usually the least expensive and are recommended for applications where impedance levels are relatively low, closed-loop gain is low, and speed requirements are moderate. BIFET inputs provide lower input-bias-currents and better bandwidth than standard bipolar inputs, but input voltage offsets and noise are generally better for the bipolar input amplifiers.

The "High Accuracy" category presents the best amplifiers for high-gain applications. A combination of low input-offset-voltage, high open-loop gain, and high CMRR provide excellent DC accuracy even at high closed-loop gain. The OP-27/37 are best for minimum input

voltage noise. The OP-08/12 provide an excellent combination of low input-bias-current, low offset voltage, and moderate power drain. The OP-07 offers a selection of the lowest offset voltages (60 $\mu$ V max to 250 $\mu$ V max) combined with low input-bias-current ( $\pm$ 2nA max to  $\pm$ 12nA max) and has become an industry standard for high-precision applications.

PMI is a leader in op amps featuring low power consumption. The OP-20/21/22 are micropower op amps that operate with only a few microamps of supply drain. PSRR and CMRR are high, and the input-voltage-range is wide. Such features work together to make these amplifiers ideal for battery-powered applications or for operation from a single supply voltage. The OP-22 can be programmed to operate over any supply current from 1 $\mu$ A to 400 $\mu$ A and is excellent for battery-powered designs.

	General Purpose						Precision High Speed BIFET Input	High Accuracy				Low Power Low Input Bias Current	Micropower										
	Bipolar			BIFET				OP-15	OP-16, OP-17	OP-215	OP-05, OP-06, OP-07		OP-08, OP-12	OP-27, OP-37	OP-207, OP-227	PM-108	PM-2108	OP-20	OP-21	OP-22, OP-32	OP-220	OP-221	OP-420
Monolithic Technology	OP-01	OP-02	OP-04, OP-14	OP-09, OP-11	PM-155	PM-156, PM-157	OP-05, OP-06, OP-07					OP-08, OP-12											
Bipolar Input	●	●	●	●					●	●	●	●	●	●	●	●	●	●	●	●	●	●	●
BIFET					●	●	●	●															
Packages																							
Single	●	●			●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●
Dual			●						●				●					●	●				
Quad				●																●	●		
Low Offset, $V_{OS} \leq 1\text{mV}$	●			●			●	●	●	●	●	●			●	●	●	●	●	●	●	●	●
Low Bias Current																							
$I_B < 100\text{pA}$					●	●																	
$I_B < 2\text{nA}$							●	●	●		●		●	●									
High Gain, $A_{VOL} \geq 200,000$											●		●	●			●	●	●	●	●	●	●
High Slew Rate, $SR \geq 10\text{V}/\mu\text{s}$	●					●		●															
Low Power																							
$I_{SY} \leq 1\text{mA/Amplifier}$											●		●	●			●			●	●	●	●
$I_{SY} \leq 100\mu\text{A/Amplifier}$															●		●	●		●	●	●	●

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# D/A CONVERTERS

## SELECTION GUIDE

PMI offers a complete line of digital-to-analog converters (DACs), all which are guaranteed to be monotonic over their operating temperature ranges, and some which have become industry standards. The DACs have been arranged in a

matrix which highlights their primary characteristics so that the user can easily narrow the selection according to specific requirements. More detailed specifications are then tabulated in each product group.

	Voltage Output						Current Output						
	DAC-01	DAC-02, DAC-03	DAC-05	DAC-06	DAC-208	DAC-210	DAC-08	DAC-10	DAC-20	DAC-100	DAC-312	μP Comp DAC-808 DAC-888	Comanding DAC-86 DAC-88, DAC-89
Resolution													
6-Bits	•												
8-Bits					•		•		•			•	•
10-Bits		•	•	•		•		•		•			
12-Bits											•		
Input Coding													
BCD									•				
Binary							•	•			•	•	
Complementary Binary	•									•			
Sign Magnitude		•	•		•	•							
Comanding												•	•
Two's Complement				•									
Complementary Current Outputs							•	•	•		•	•	
Internal Reference	•	•	•	•	•	•				•			
Logic Threshold Control							•	•	•		•		•
JAN Qualified							•						
Operating Temp. Range													
Military	•		•	•	•	•	•	•		•	•	•	
Industrial										•	•		•
Commercial	•	•	•	•	•	•	•	•	•	•	•	•	•

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## AMP-01

LOW-NOISE

PRECISION

PRELIMINARY

INSTRUMENTATION AMPLIFIER

### FEATURES

- Low Offset Voltage .....  $15\mu\text{V}$
- Very Low Offset Voltage Drift .....  $0.1\mu\text{V}/^\circ\text{C}$
- Low Noise .....  $0.2\mu\text{V}_{\text{p-p}}$  (0.1Hz to 10Hz)
- Excellent Output Drive .....  $\pm 10\text{V}$  at  $\pm 50\text{mA}$
- Capacitive Load Stability ..... to  $1\mu\text{F}$
- Gain Range ..... 0.1 to 10,000
- Excellent Linearity ..... 16-Bit at  $G = 1000$
- High CMR ..... 140dB ( $G = 1000$ )
- Low Bias Current ..... 1nA
- May be Configured as a Precision Op-Amp
- Output-Stage Thermal Shutdown

### GENERAL DESCRIPTION

The AMP-01 is a monolithic instrumentation amplifier designed for high-precision data acquisition and general

instrumentation applications. The design combines the conventional features of an instrumentation amplifier with a high-current output stage. The output remains stable with high capacitance loads ( $1\mu\text{F}$ ), a unique ability for an instrumentation amplifier. Consequently, the AMP-01 can amplify low-level signals for transmission through long cables without requiring an output buffer. The output stage may be configured as a voltage or current generator.

Input offset voltage is very low ( $15\mu\text{V}$ ) which generally eliminates the need for an external null potentiometer. Temperature changes have minimal effect on offset;  $\text{TCV}_{\text{IOS}}$  is typically  $0.1\mu\text{V}/^\circ\text{C}$ . Excellent low-frequency noise performance is achieved with a minimal compromise on input protection. Bias current is very low, less than 10nA over the industrial temperature range.

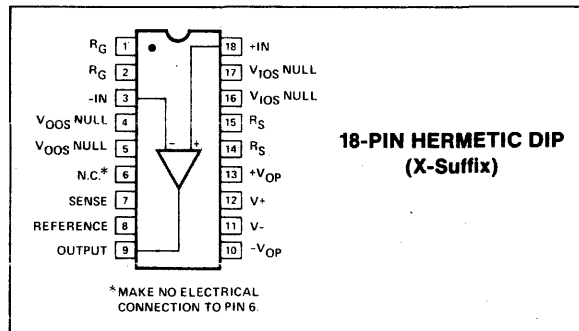
### ORDERING INFORMATION†

PACKAGE	OPERATING TEMPERATURE RANGE
CERDIP 18-PIN	
AMP01AX*	MIL
AMP01BX*	MIL
AMP01EX	IND
AMP01FX	IND

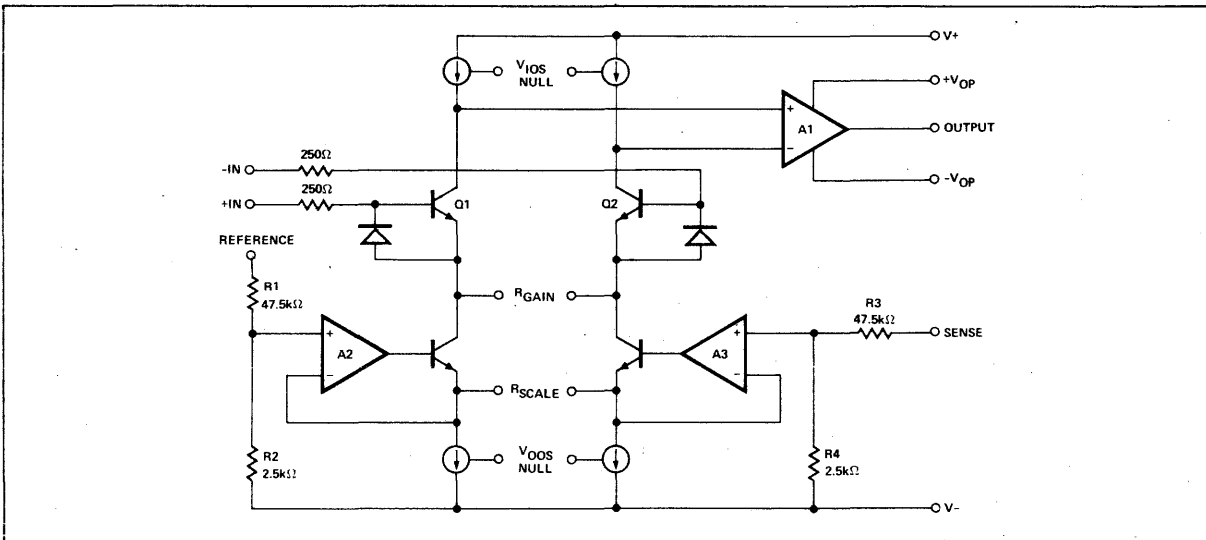
\*Also available with MIL-STD-883B Processing. To order add /983 as a suffix to the part number. Screening Procedure: 1984 Data Book, Section 3.

†All commercial and industrial temperature range parts are available with burn-in per MIL-STD-883. Ordering Information: 1984 Data Book, Section 2.

### PIN CONNECTIONS



### SIMPLIFIED SCHEMATIC



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LINEAR

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**LOW-NOISE, MATCHED  
DUAL MONOLITHIC  
TRANSISTOR**

**MAT-02  
PRELIMINARY**

**FEATURES**

- **Low Offset Voltage** ..... **50 $\mu$ V Max**
- **Low Noise Voltage at 100Hz** ..... **1.8nV/ $\sqrt$ Hz Typ**
- **High Gain ( $h_{FE}$ )** ..... **500 Min at  $I_C = 1mA$**
- ..... **300 Min at  $I_C = 1\mu A$**
- **Excellent Log Conformance** .....  **$r_{BE} \approx 0.3\Omega$**

**ORDERING INFORMATION**

$T_A = 25^\circ C$ $V_{OS} \text{ Max}$ ( $\mu V$ )	PACKAGE	OPERATING TEMPERATURE RANGE
50	MAT02AH*	MIL
50	MAT02EH	IND
150	MAT02BH*	MIL
150	MAT02FH	IND

\*Also available with MIL-STD-883B Processing. To order add /883 as a suffix to the part number. Screening Procedure: 1984 Data Book, Section 3.  
†All commercial and industrial temperature range parts are available with burn-in per MIL-STD-883. Ordering Information: 1984 Data Book, Section 2.

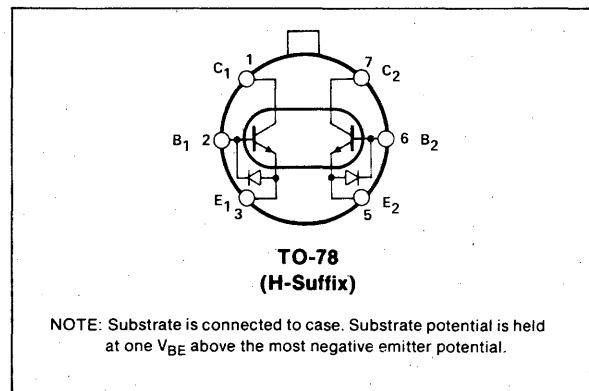
**GENERAL DESCRIPTION**

The design of the MAT-02 series of NPN dual monolithic transistors is optimized for very low noise, low drift, and low  $r_{BE}$ . Precision Monolithics' exclusive Silicon Nitride "Triple-Passivation" process stabilizes the critical device parameters over wide ranges of temperature and elapsed time. Also, the high current gain ( $h_{FE}$ ) of the MAT-02 is maintained over a wide range of collector current. Exceptional characteristics of the MAT-02 include offset voltage of 50 $\mu$ V max (A/E grades) and 150 $\mu$ V max (B/F grades). Device performance is specified over the full military temperature range as well as at 25 $^\circ C$ .

Input protection diodes are provided across the emitter-base junctions to prevent degradation of the device characteristics due to reverse-biased emitter current. The substrate is clamped to the most negative emitter by the parasitic isolation junction created by the protection diodes. This results in isolation between the transistors.

The MAT-02 should be used in any application where low noise is a priority. The MAT-02 can be used as an input stage to make an amplifier with noise voltage of less than 2.0nV/ $\sqrt$ Hz at 100Hz. Other applications, such as log/anti-log circuits may use the excellent logging conformity of the MAT-02. Typical excess emitter resistance is only 0.3 $\Omega$  to 0.4 $\Omega$ . The MAT-02 electrical characteristics approach those of an ideal transistor when operated over a collector current range of 1 $\mu$ A to 10mA.

**PIN CONNECTIONS**



**ELECTRICAL CHARACTERISTICS** at  $V_{CB} = 15V$ ,  $I_C = 10\mu A$ ,  $T_A = 25^\circ C$ , unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	MAT-02A/E			MAT-02B/F			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
Offset Voltage	$V_{OS}$	$I_C = 10\mu A$	—	—	50	—	—	150	$\mu V$
Bias Current	$I_B$	$I_C = 10\mu A$	—	—	33	—	—	50	nA
Offset Current	$I_{OS}$	$I_C = 10\mu A$	—	—	0.6	—	—	2	nA
Current Gain	$h_{FE}$	$I_C = 1mA$	500	—	—	400	—	—	
		$I_C = 100\mu A$	500	—	—	400	—	—	
		$I_C = 10\mu A$	400	—	—	300	—	—	
		$I_C = 1\mu A$	300	—	—	200	—	—	
Current Gain Match	$h_{FE} \text{ Match}$	$10\mu A \leq I_C \leq 1mA$	—	—	2	—	—	4	%
Excess Emitter Resistance	$r_{BE}$		—	0.3	0.5	—	0.3	0.5	$\Omega$
Noise Voltage	$e_n$	$I_C = 100\mu A$ , $f = 100Hz$	—	1.8	—	—	1.8	—	nV/ $\sqrt$ Hz
Collector Base Voltage	$V_{CB}$		—	—	40	—	—	40	V

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## OP-32

HIGH-SPEED

PROGRAMMABLE MICROPOWER

PRELIMINARY

OPERATIONAL AMPLIFIER (SINGLE OR DUAL SUPPLY)

### FEATURES

- Programmable Supply Current .....  $1\mu\text{A}$  to  $2\text{mA}$
- Single Supply Operation .....  $+3\text{V}$  to  $+30\text{V}$
- Dual Supply Operation .....  $\pm 1.5\text{V}$  to  $\pm 15\text{V}$
- Low Input Offset Voltage .....  $100\mu\text{V}$
- Low Input Offset Voltage Drift .....  $0.75\mu\text{V}/^\circ\text{C}$
- High Common-Mode Input Range ...  $V^-$  to  $V^+$  ( $-1.5\text{V}$ )
- High CMRR and PSRR .....  $115\text{dB}$
- High Open-Loop Gain .....  $2000\text{V}/\text{mV}$
- $\pm 30\text{V}$  Input Overvoltage Protection
- Fast .....  $1\text{V}/\mu\text{s}$  @  $I_{\text{SET}} = 20\mu\text{A}$
- LM4250 Pinout
- Compensated for Minimum Gain of 10

voltage, and set current. High CMRR and PSRR ensure precision performance when the OP-32 is used with an unregulated battery or vehicular electrical systems.

The wide input voltage range, including the negative supply or ground, allows use in single-battery applications. The OP-32 is characterized over a wide supply range of  $\pm 1.5\text{V}$  to  $\pm 15\text{V}$ . This guarantees predictable performance with any commonly available supply.

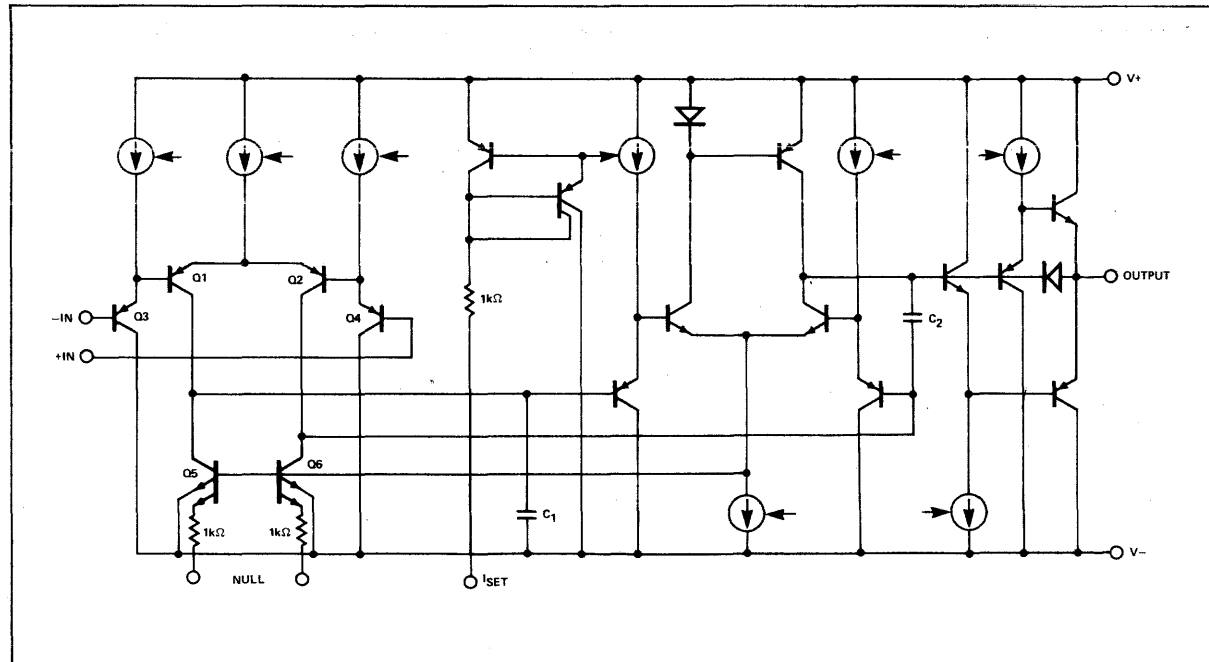
The ability to operate at relatively high speed with low power consumption makes this amplifier ideal for remote applications where power is limited. The programmability allows each amplifier in a system to be set for the minimum power consumption necessary for each specific application. Programmability also makes it possible to adjust the bandwidth and phase shift.

### GENERAL DESCRIPTION

The OP-32 is a high-speed, high-gain programmable operational amplifier. Both offset voltage and offset current are low, and both are stable with changes in temperature, supply

The OP-32 pinout is identical to the LM4250 and many other micropower operational amplifiers. This allows easy upgrading of system performance.

### SIMPLIFIED SCHEMATIC



The premium performance of this product is achieved through an advanced processing technology. All Precision Monolithics products are guaranteed to meet or exceed published specifications.

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## OP-32 HIGH-SPEED PROGRAMMABLE MICROPOWER OPERATIONAL AMPLIFIER — PRELIMINARY

### ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	±18V
Power Dissipation	500mW
Differential Input Voltage	±30
Input Voltage	Supply Voltage
Storage Temperature Range	
Z Package	-65°C to +150°C
P Package	-55°C to +125°C

### Operating Temperature Range

OP-32A, B (Z package)	-55°C to +125°C
OP-32E, F & G (Z or P package)	-25°C to +85°C
Lead Temperature Range (Soldering, 60 sec)	300°C
DICE Junction Temperature	-65°C to +150°C

### NOTE:

1. Absolute ratings apply to both DICE and packaged parts, unless otherwise noted.

### ELECTRICAL CHARACTERISTICS at $V_S = \pm 1.5V$ to $\pm 15V$ , $1\mu A \leq I_{SET} \leq 30\mu A$ , $T_A = +25^\circ C$ , unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-32A/E			OP-32B/F			OP-32G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	$V_{OS}$		—	100	300	—	200	500	—	400	1000	$\mu V$
Input Offset current	$I_{OS}$	$V_{CM} = 0$	—	0.2	1	—	0.3	2	—	0.5	3	nA
Input Bias Current (Note 1)	$I_B$	$I_{SET} = 1\mu A$	—	2.6	5	—	3.0	7.5	—	4.0	10	nA
		$I_{SET} = 10\mu A$	—	19	30	—	24	35	—	30	50	
		$I_{SET} = 30\mu A$	—	45	75	—	50	100	—	60	125	
Input Voltage Range	IVR	$V_S = \pm 15V$	-15.0/13.5	—	—	-15.0/13.5	—	—	-15.0/13.5	—	—	V
Common-Mode Rejection Ratio	CMRR	$V_S = \pm 15V$ $-15V \leq V_{CM} \leq +13.5V$	100	115	—	95	110	—	85	100	—	dB
Power Supply Rejection Ratio	PSRR	$V_S = \pm 1.5V$ to $\pm 15V$ ; and $V_- = 0V$ , $V_+ = 3V$ to $30V$ .	—	1	6	—	3	12	—	10	25	$\mu V/V$
Large-Signal Voltage Gain	$A_{VO}$	$V_S = \pm 15V$ , $R_L = 100k\Omega$ $R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	1000	2000	—	750	1500	—	500	1000	—	V/mV
Output Voltage Swing	$V_O$	$V_S = \pm 1.5V$ $R_L = 100k\Omega$ , $I_{SET} = 1\mu A$ $R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	$\pm 0.8 \pm 0.88$		—	$\pm 0.8 \pm 0.88$		—	$\pm 0.75 \pm 0.85$		—	V
		$V_S = \pm 15V$ $R_L = 100k\Omega$ , $I_{SET} = 1\mu A$ $R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	$\pm 14 \pm 14.2$		—	$\pm 14 \pm 14.2$		—	$\pm 13.8 \pm 14.2$		—	V
Gain-Bandwidth Product		$I_{SET} = 1\mu A$ , $R_L = 100k\Omega$	—	100	—	—	100	—	—	100	—	kHz
		$I_{SET} = 30\mu A$ , $R_L = 10k\Omega$	—	4500	—	—	4500	—	—	4500	—	
Slew Rate	SR	$V_S = \pm 15V$ , $I_{SET} = 10\mu A$ , $R_L = 10k\Omega$	—	0.6	—	—	0.6	—	—	0.6	—	$V/\mu s$
Supply Current No Load	$I_{SY}$	$V_S = \pm 15V$ , $I_{SET} = 1\mu A$	—	15	17	—	15	19	—	15	21	$\mu A$
		$I_{SET} = 10\mu A$	—	150	170	—	150	190	—	150	200	
		$I_{SET} = 30\mu A$	—	450	525	—	450	600	—	450	650	
		$V_S = \pm 1.5V$ , $I_{SET} = 1\mu A$	—	10.5	12.5	—	11	15	—	11	18	
		$I_{SET} = 10\mu A$	—	105	125	—	110	150	—	110	180	
		$I_{SET} = 30\mu A$	—	350	400	—	350	450	—	350	500	

### NOTE:

1.  $I_B$  and  $I_{OS}$  are measured at  $V_{CM} = 0$ .

## OP-32 HIGH-SPEED PROGRAMMABLE MICROPOWER OPERATIONAL AMPLIFIER — PRELIMINARY

**ELECTRICAL CHARACTERISTICS** at  $V_S = \pm 1.5V$  to  $\pm 15V$ ,  $1\mu A \leq I_{SET} \leq 30\mu A$ ,  $-55^\circ C \leq T_A \leq +125^\circ C$  for OP-32AZ and OP-32BZ, unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-32A			OP-32B			UNITS
			MIN.	TYP	MAX	MIN	TYP	MAX	
Average Input Offset Voltage Drift (Note 1)	$TCV_{OS}$	Unnullified	—	0.75	2.0	—	1.0	2.0	$\mu V/^\circ C$
Input Offset Voltage	$V_{OS}$		—	175	400	—	350	600	$\mu V$
Input Offset Current	$I_{OS}$	$V_{CM} = 0$	—	0.2	1	—	0.3	2	nA
Average Input Offset Current Drift	$TCI_{OS}$	(Note 1)	—	1	10	—	3	15	$\mu A/^\circ C$
Input Bias Current (Note 2)	$I_B$	$I_{SET} = 1\mu A$	—	2.8	5	—	3.3	7.5	nA
		$I_{SET} = 10\mu A$	—	21	30	—	27	35	
		$I_{SET} = 30\mu A$	—	40	75	—	50	100	
Input Voltage Range	IVR	$V_S = \pm 15V$	-15.0/13.5	—	—	-15.0/13.5	—	—	V
Common-Mode Rejection Ratio	CMRR	$V_S = \pm 15V$ $-15V \leq V_{CM} \leq +13.5V$	90	110	—	86	105	—	dB
Power Supply Rejection Ratio	PSRR	$V_S = \pm 1.5V$ to $\pm 15V$ & $V_- = 0V$ , $V_+ = 3V$ to $30V$	—	2	10	—	2.5	20	$\mu V/V$
Large-Signal Voltage Gain	$A_{VO}$	$V_S = \pm 15V$ $R_L = 100k\Omega$	200	400	—	200	500	—	V/mV
		$R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	500	1000	—	300	750	—	
Output Voltage Swing	$V_O$	$V_S = \pm 1.5V$ $R_L = 100k\Omega$ , $I_{SET} = 1\mu A$	$\pm 0.65$	$\pm 0.75$	—	$\pm 0.65$	$\pm 0.75$	—	V
		$R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$							
		$V_S = \pm 15V$ $R_L = 100k\Omega$ , $I_{SET} = 1\mu A$	$\pm 13.6$	$\pm 14.0$	—	$\pm 13.0$	$\pm 13.5$	—	V
Supply Current No Load	$I_{SY}$	$V_S = \pm 15V$ , $I_{SET} = 1\mu A$	—	16	18	—	16	20	$\mu A$
		$I_{SET} = 10\mu A$	—	160	180	—	160	200	
		$I_{SET} = 30\mu A$	—	450	550	—	450	600	
		$V_S = \pm 1.5V$ , $I_{SET} = 1\mu A$	—	12	14	—	12	17	
		$I_{SET} = 10\mu A$	—	120	140	—	120	170	
		$I_{SET} = 30\mu A$	—	360	450	—	360	500	

**NOTE:**

1. Sample tested.
2.  $I_B$  and  $I_{OS}$  are measured at  $V_{CM} = 0$ .

LINEAR  
Precision Monolithics

## OP-32 HIGH-SPEED PROGRAMMABLE MICROPOWER OPERATIONAL AMPLIFIER — PRELIMINARY

**ELECTRICAL CHARACTERISTICS** at  $V_S = \pm 1.5V$  to  $\pm 15V$ ,  $1\mu A \leq I_{SET} \leq 30\mu A$ ,  $-25^\circ C \leq T_A \leq +85^\circ C$  for OP-32EP/EZ, OP-32FP/FZ, OP-32GP and OP-32GZ, unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-32E			OP-32F			OP-32G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Average Input Offset Voltage Drift (Note 1)	$TCV_{OS}$	Unnull'd	—	0.75	1.5	—	1.0	2.0	—	1.5	3.0	$\mu V/^\circ C$
Input Offset Voltage	$V_{OS}$		—	100	300	—	200	600	—	500	1200	$\mu V$
Input Offset Current	$I_{OS}$	$V_{CM} = 0$	—	0.2	1	—	0.3	2	—	0.5	3	nA
Average Input Offset Current Drift	$TCI_{OS}$	(Note 1)	—	2	10	—	3	15	—	5	25	$pA/^\circ C$
Input Bias Current (Note 2)	$I_B$	$I_{SET} = 1\mu A$	—	2.6	5	—	3.0	7.5	—	4.0	10	nA
		$I_{SET} = 10\mu A$	—	19	30	—	24	35	—	30	50	
		$I_{SET} = 30\mu A$	—	45	75	—	50	100	—	60	125	
Input Voltage Range	IVR	$V_S = \pm 15V$	-15.0/13.5	—	—	-15.0/13.5	—	—	-15.0/13.5	—	—	V
Common-Mode Rejection Ratio	CMRR	$V_S = \pm 15V$ & $-15V \leq V_{CM} \leq +13.5V$	95	110	—	90	105	—	80	100	—	dB
Power Supply Rejection Ratio	PSRR	$V_S = \pm 1.5V$ to $\pm 15V$ & $V_- = 0V$ , $V_+ = 3V$ to $30V$	—	3.2	10	—	10	32	—	32	56	$\mu V/V$
Large-Signal Voltage Gain	$A_{VO}$	$V_S = \pm 15V$ $R_L = 100k\Omega$	750	1000	—	500	1200	—	400	1000	—	V/mV
		$R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	750	1000	—	500	1200	—	400	1000	—	
Output Voltage Swing	$V_O$	$V_S = \pm 1.5V$ $R_L = 100k\Omega$ , $I_{SET} = 1\mu A$ $R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	$\pm 0.70$	$\pm 0.75$	—	$\pm 0.65$	$\pm 0.75$	—	$\pm 0.6$	$\pm 0.7$	—	V
		$V_S = \pm 15V$ $R_L = 100k\Omega$ , $I_{SET} = 1\mu A$ $R_L = 10k\Omega$ , $10\mu A \leq I_{SET} \leq 30\mu A$	$\pm 13.8$	$\pm 14.1$	—	$\pm 13.6$	$\pm 14.1$	—	$\pm 13.0$	$\pm 14.0$	—	V
Supply Current No Load	$I_{SY}$	$V_S = \pm 15V$ , $I_{SET} = 1\mu A$	—	16	18	—	16	20	—	16	25	$\mu A$
		$I_{SET} = 10\mu A$	—	160	180	—	160	200	—	160	250	
		$I_{SET} = 30\mu A$	—	450	550	—	450	600	—	450	650	
		$V_S = \pm 1.5V$ , $I_{SET} = 1\mu A$	—	12	14	—	12	17	—	12	25	
		$I_{SET} = 10\mu A$	—	120	140	—	120	170	—	120	200	
		$I_{SET} = 30\mu A$	—	360	450	—	360	500	—	360	550	

**NOTE:**

1. Sample tested.
2.  $I_B$  and  $I_{OS}$  are measured at  $V_{CM} = 0$ .

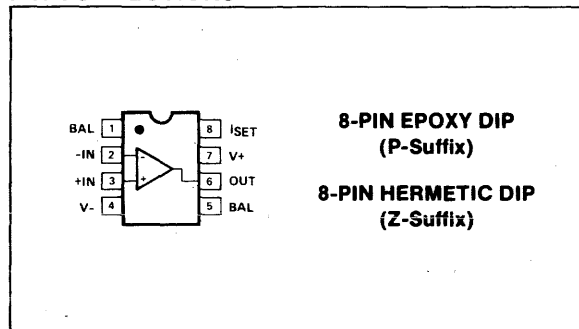
### ORDERING INFORMATION†

$T_A = 25^\circ C$ $V_{OS}$ MAX ( $\mu V$ )	PACKAGE		OPERATING TEMPERATURE RANGE
	EPOXY DIP 8-PIN	HERMETIC DIP 8-PIN	
300	OP32AP	OP32AZ*	MIL
300	OP32EP	OP32EZ	IND
500	OP32BP	OP32BZ*	MIL
500	OP32FP	OP32FZ	IND
1000	OP32GP	OP32GZ	IND

\*Also available with MIL-STD-883B Processing. To order add /883 as a suffix to the part number. Screening Procedure: 1984 Data Book, Section 3.

†All commercial and industrial temperature range parts are available with burn-in per MIL-STD-883. Ordering Information: 1984 Data Book, Section 2.

### PIN CONNECTIONS



## SALES OFFICES

### Headquarters, Factory and Regional Sales Offices

#### Santa Clara

1500 Space Park Dr.  
Santa Clara, CA 95050  
(408) 727-6616  
TWX 910-338-2102

#### Los Angeles

6033 W. Century Blvd., Suite 595  
Los Angeles, CA 90045  
(213) 642-0124  
TWX 910-328-6591

#### Chicago

1325 Remington Rd., Suite "D"  
Schaumburg, IL 60195  
(312) 885-8440, (800) 323-8755  
TWX 910-222-1808

#### Dallas

11325 Pegasus St., Suite W-126  
Dallas, TX 75238  
(214) 341-1742, (800) 323-8755  
TWX 910-861-4079

#### Boston

209 W. Central St.  
Natick, MA 01760  
(617) 655-8900  
TWX 710-386-0114

### Europe

#### Headquarters

Bourns/PMI  
Zugerstrasse 74  
6340 Baar  
Switzerland  
Phone: 042-33 33 33  
Telex: 868 722

#### Benelux

Bourns (Nederland) B.V.  
Van Tuyl van Serooskerkestr. 85  
P.O. Box 37  
2270 AA Voorburg  
Phone: 070-87 44 00  
Telex: 32 023

#### France

Ohmic SA  
21/23 rue des Ardennes  
75019 Paris  
Phone: 01-203 96 33  
Telex: 230 008

#### Germany

Bourns GmbH  
Postfach 1155  
Eberhardstrasse 63  
7000 Stuttgart 1  
Phone: 0711-24 29 36  
Telex: 721 556

#### Switzerland

Bourns (Schweiz) AG  
Zugerstrasse 74  
6340 Baar  
Phone: 042-33 33 33  
Telex: 868 722

#### United Kingdom

Bourns Electronics Ltd.  
Hodford House  
17/27 High Street  
Hounslow, Middlesex TW3 1TE  
Phone: 01-572 65 31  
Telex: 264 485

### Japan

Nippon PMI Corporation  
Haratetsu Building  
4-1-11, Kudan Kita  
Chiyoda-ku, Tokyo  
Phone: (03) 234-1141  
Telex: 781 J 27632

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# Precision Monolithics Incorporated





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Circuits Magazine**



# Inverting Switching Regulator

4391

## Features

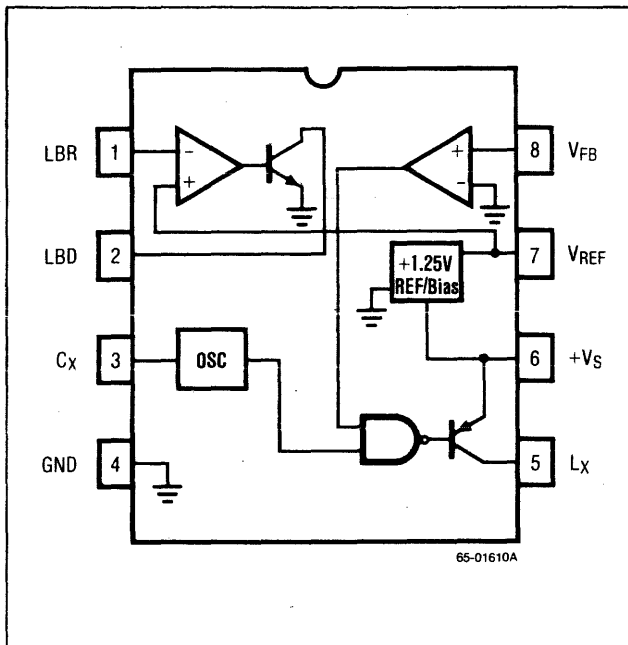
- Converts a Positive Voltage into a Negative Voltage
- Specifically Designed for Low Power Applications, Including Batteries
- Adjustable Output Voltage
- High Switch Current Capability
- Low Quiescent Supply Current — 175 $\mu$ A Typical
- Eight Pin Mini Package
- Low Battery Detection Circuitry

## Description

Raytheon's micro-power inverting switching regulator, RM/RC4391, is a monolithic low power switching regulator specifically designed for low power inverting applications. The RC4391 contains an internal 1.25 Volt bandgap voltage reference, switch transistor, comparator, free running oscillator and low battery detection circuitry. These components are interconnected to minimize the number of external components required in typical inverting applications (see Figure 2). The RC4391 requires an inductor, diode, timing capacitor, and an R<sub>2</sub>, R<sub>1</sub> network to achieve a negative output voltage. The RC4391 allows the designer the flexibility in designing unconventional applications such as replacing the internal bandgap reference with an external or system reference, or using the low battery detection comparator and transistor as voltage level detectors or for signal generation.

A typical application would combine the RC4391 with the RC4193 micro-power switching regulator to convert a single input voltage into a  $\pm 12$  Volt or  $\pm 15$  Volt power supply. The single voltage can be either from a battery, bridge rectifier or existing +5.0 Volt line on card for this application (see Figure 3).

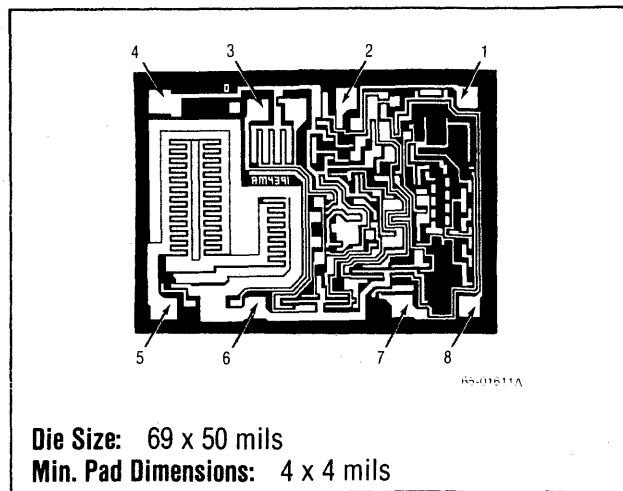
## Functional Block Diagram



## Thermal Characteristics

	8-Lead Plastic DIP	8-Lead Ceramic DIP
Max. Junction Temp.	125°C	175°C
Max. P <sub>D</sub> T <sub>A</sub> < 50°C	468mW	833mW
Therm. Res. $\theta_{JC}$		45°C/W
Therm. Res. $\theta_{JA}$	160°C/W	150°C/W
For T <sub>A</sub> > 50°C Derate at	6.25mW per °C	8.33mW per °C

## Mask Pattern



**Electrical Characteristics** (Circuit of Figure 2.,  $V_{IN} = 6.0V$ ,  $T_A = +25^\circ C$  unless otherwise noted)

Parameter	Symbol	Condition	Min	Typ	Max	Units
Supply Voltage	$V_{IN}$		3.0		$30 -  V_{OUT} $	Volts
Supply Current	$I_{IN}$	$V_S = 4.0V$ , No External Loads		170	250	$\mu A$
		$V_S = 25V$ , No External Loads		300	500	
Output Voltage	$V_{OUT}$	$V_{OUT\ nom} = -5.0V$	-5.35	-5.0	-4.65	Volts
		$V_{OUT\ nom} = -15V$	-15.85	-15	-14.15	
Line Regulation		$V_{OUT\ nom} = -5.0V$ , $C_X = 150pF$ , $V_{IN} = 5.8V$ to $20V$		1.5	3.0	% $V_{OUT}$
		$V_{OUT\ nom} = -15V$ , $C_X = 150pF$ , $V_{IN} = 5.8V$ to $15V$		1.0	2.0	
Load Regulation		$V_{OUT\ nom} = -5.0V$ , $C_X = 350pF$ , $V_{IN} = 4.5V$ , $P_{LOAD} = 0mW$ to $75mW$		0.2	0.4	% $V_{OUT}$
		$V_{OUT\ nom} = -15V$ , $C_X = 350pF$ , $V_{IN} = 4.5V$ , $P_{LOAD} = 0mW$ to $75mW$		0.07	0.14	
Reference Voltage	$V_{REF}$		1.18	1.25	1.32	Volts
Switch Current	$I_{SW}$	Pin 5 = 5.5V	75			mA
Switch Leakage Current	$I_{CO}$	Pin 5 = -24V		0.01	5.0	$\mu A$
Timing Pin Current	$I_{CX}$	Pin 3 = 0V	10	12	14	$\mu A$
LBD Leakage Current		Pin 1 = 1.5V, Pin 2 = 6.0V		0.01	5.0	$\mu A$
LBD on Current		Pin 1 = 1.1V, Pin 2 = 0.4V	210	600		$\mu A$
LBR Bias Current		Pin 1 = 1.5V		0.7		$\mu A$

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## Electrical Characteristics

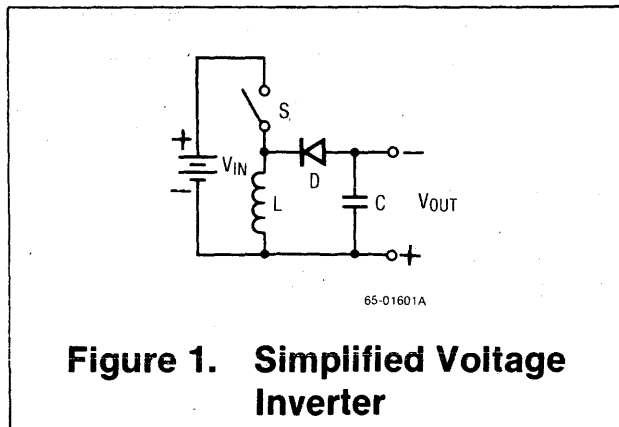
(Circuit of Figure 2.,  $V_{IN} = 6.0V$ , over the full operating temperature range unless otherwise noted)

Parameter	Symbol	Condition	Min	Typ	Max	Units
Supply Voltage	$V_{IN}$		4.0		$30 -  V_{OUT} $	Volts
Supply Current	$I_{IN}$	$V_{IN} = 25V$		300	500	$\mu A$
Reference Voltage	$V_{REF}$		1.13	1.25	1.36	Volts
Output Voltage	$V_{OUT}$	$V_{OUT\ nom} = -5.0V$	-5.5	-5.0	-4.5	Volts
		$V_{OUT\ nom} = -15V$	-16.5	-15	-13.5	
Line Regulation		$V_{OUT\ nom} = -5.0V$ , $C_X = 150pF$ , $V_{IN} = 5.8V$ to $20V$		2.0	4.0	% $V_{OUT}$
		$V_{OUT\ nom} = -15V$ , $C_X = 150pF$ , $V_{IN} = 5.8V$ to $15V$		1.5	3.0	
Load Regulation		$V_{OUT\ nom} = -5.0V$ , $C_X = 350pF$ , $V_{IN} = 4.5V$ , $P_{LOAD} = 0mW$ to $75mW$		0.2	0.5	mV
		$V_{OUT\ nom} = -15V$ , $C_X = 350pF$ , $V_{IN} = 4.5V$ , $P_{LOAD} = 0mW$ to $75mW$		0.15	0.3	
Switch Leakage Current	$I_{CO}$	Pin 5 = $-24V$		0.1	30	$\mu A$

## Principles of Operation

The basic switching inverter circuit is the building block on which the complete inverting application is based.

A simplified diagram of the voltage inverter circuit with ideal components and no feedback circuitry is shown in (Figure 1). When the switch S is closed, charging current from the



**Figure 1. Simplified Voltage Inverter**

battery flows through the inductor  $L$ , which builds up a magnetic field, increasing as the switch is held closed. When the switch is opened, the magnetic field collapses, and the energy stored in the magnetic field is converted into a current which flows through the inductor in the same direction as the charging current. Because there is no path for this current to flow through the switch, the current must flow through the diode to charge the capacitor  $C$ . The key to the inversion is the ability of the inductor to become a source when the charging current is removed.

The equation  $V = (L) (di/dt)$  gives the maximum possible voltage across the inductor; in the actual application, feedback circuitry and the output capacitor will decrease the output voltage to a regulated fixed value.

A complete schematic for the standard inverting application is shown in (Figure 3). The ideal switch in the simplified diagram is replaced by

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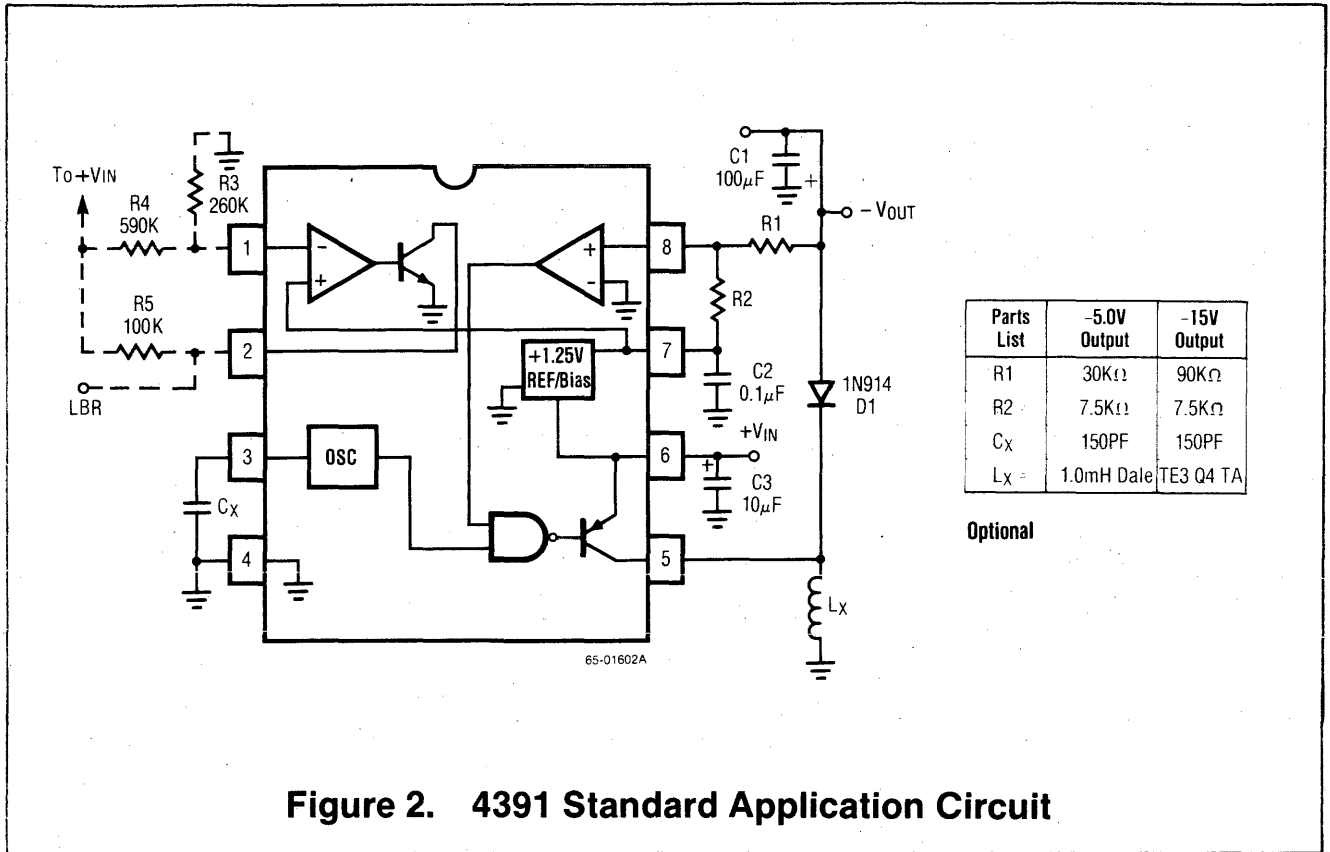


Figure 2. 4391 Standard Application Circuit

the PNP transistor switch between pins 5 and 6. C1 functions as the output filter capacitor, and D1 and L<sub>x</sub> replace D and L.

When power is first applied, the ground sensing comparator (Pin 8) compares the output voltage to the +1.25V voltage reference. Because C1 is initially discharged a positive voltage is applied to the comparator, and the output of the comparator gates the square wave oscillator. This gated square wave signal turns on, then off, the PNP output transistor. This turning on and off of the output transistor performs the same function as opening and closing the ideal switch in the simplified diagram; i.e., it stores energy in the inductor during the on time and releases it into the capacitor during the off time.

The comparator will continue to gate the oscillator to the switch transistor until enough energy has been stored in the output capacitor to make the comparator input voltage decrease to less than zero volts. The voltage applied to the comparator is set by the output voltage, the reference voltage, and the ratio of R1 to R2. To adjust the output voltage, set R2 to about 10KΩ

and change R1 according to the following equation:

$$R1 = \frac{(V_{OUT}) (R2)}{V_{REF}}$$

Efficiency and load regulation will improve if a quality high Q inductor is used. A ferrite core toroid inductor is recommended. The inductor value and timing capacitor value must be carefully tailored to the ripple, input voltage range, output voltage, switch transistor maximum current, and output load current requirements of the application. The following equation will help select component values for high output current at input voltages less than or equal to the output voltage.

$$I_{OUT} (average) \approx \frac{(V_{IN} - V_{SW})^2}{(L) (8.0) (f_O) (V_{OUT} + V_D)}$$

- where L = Inductor Value in Henrys
- f<sub>O</sub> = Switching Frequency
- V<sub>SW</sub> = Saturation Voltage of Switching Transistor
- V<sub>D</sub> = Diode Voltage

The input voltage is a squared term; if the input voltage decreases, even by a small amount, the output current capability will decrease sharply. Since the L term is in the denominator, lower inductor values will increase the output current capability (until the inductor current rises so high that the inductor saturates — an air gapped core can help this). Lowering the switching frequency will also help, but will increase the ripple. Also, the frequency should not be so low that the inductor current rises too high and saturates the inductor or exceeds the maximum current rating of the switch transistor, which is 375 milliamps.

For  $V_{IN} < |V_{OUT}|$

$$I_{MAX} \approx \frac{(V_{IN} - V_{SW})^2}{2 L f_O}$$

For  $V_{IN} > |V_{OUT}|$

$$I_{MAX} \approx 2 \times I_{LOAD}$$

The value of the timing capacitor is set according to the following equation:

$$f_{OP} \text{ (Hz)} = \frac{2.14 \times 10^{-6}}{C_x}$$

The square wave output of the oscillator is internal and cannot be directly measured, but is equal in frequency to the triangle waveform measurable at pin 3. The switch transistor is normally on when the triangle waveform is ramping up and off when ramping down. Capacitor selection depends on the application; higher operating frequencies will reduce the output voltage ripple while trading off efficiency and load regulation. Keep the capacitor lead length short.

The low battery detector provides a method of signaling a computer system or an LCD display when the battery voltage decays below a set voltage. This voltage is predetermined by a voltage divider tied to the battery input and by the +1.25V voltage reference:

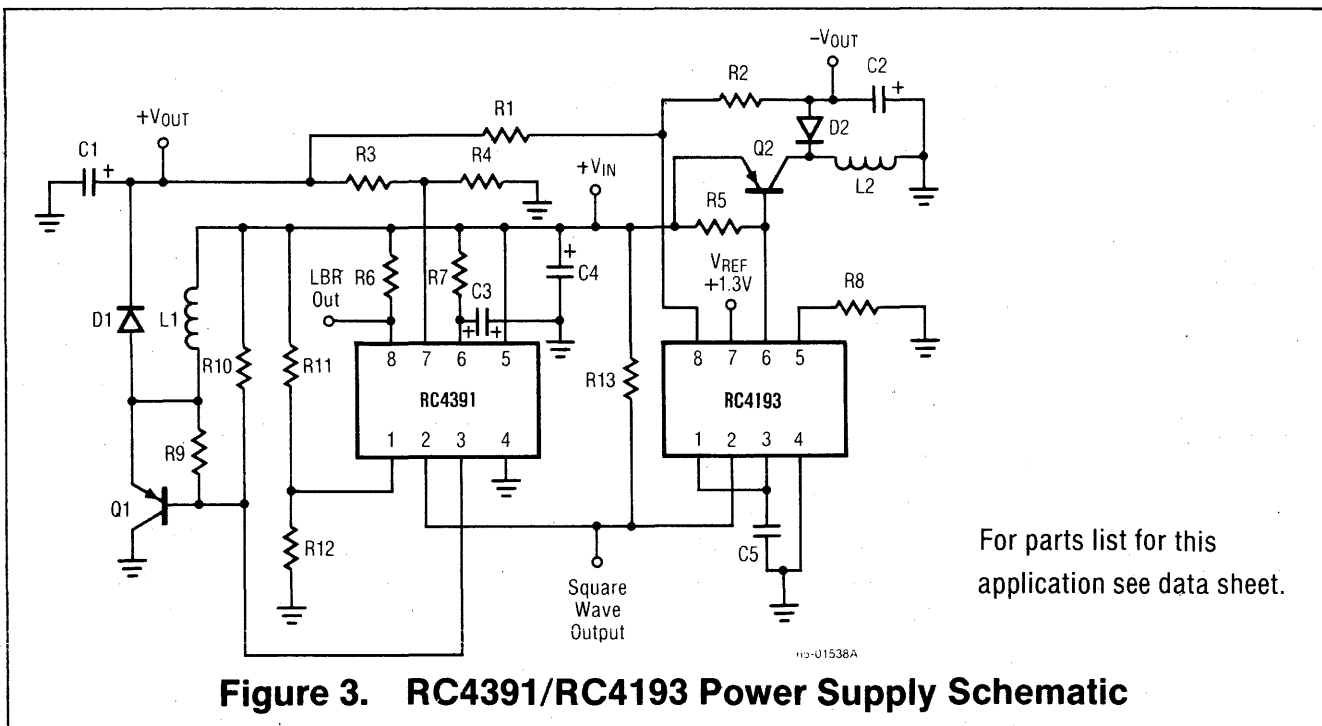
$$V_{TH} = \left( \frac{R_4}{R_3} + 1.0 \right) V_{REF}$$

The open collector output transistor must have an external load resistor selected according to the input drive requirement of the logic family used.

An inverting and non-inverting power supply circuit using a step up RC4193 and an inverting 4391 shown in (Figure 3). The component

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values given will provide a positive 15 Volt output and a negative 15 Volt output. These voltages can be adjusted by changing the value of R1,  $V_{OUT} = (R1/R2 + 1.0) (V_{REF})$ , but note that the battery voltage must be less than the programmed positive output voltage. The reference for the negative output voltage is derived from the positive output voltage so that the output voltages will track with changes in temperature and battery voltage.

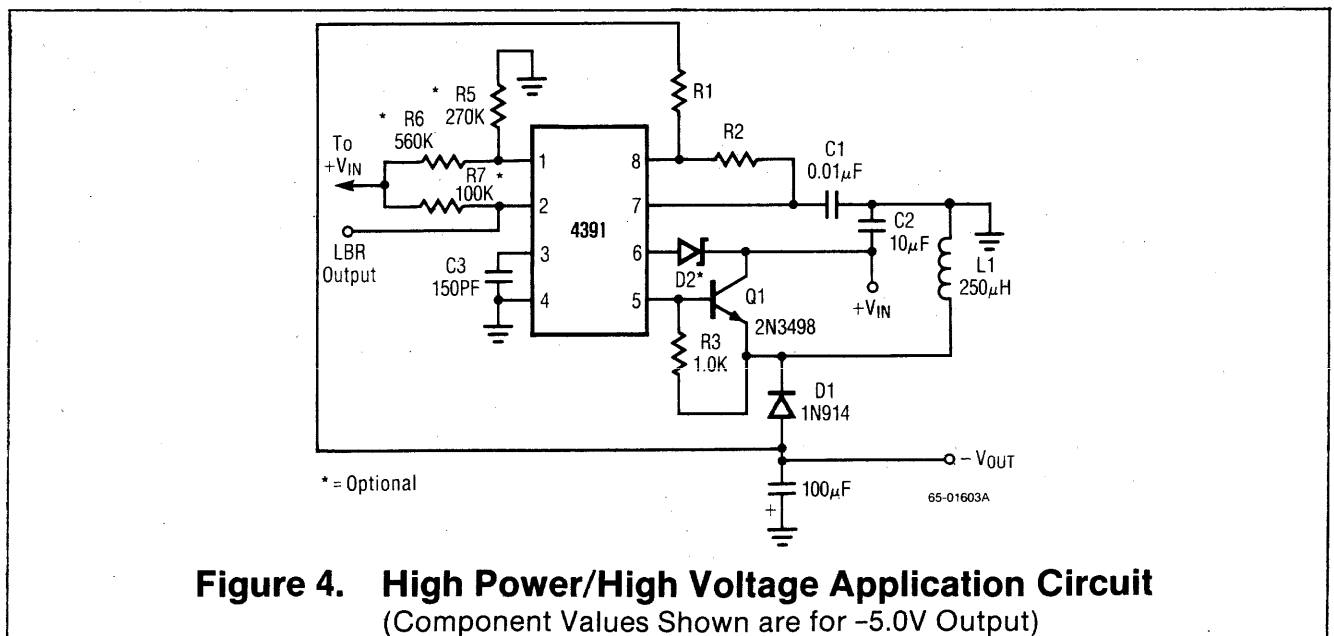
The inductor and timing capacitor values shown were chosen for best operation with a 12 Volt nominal battery input that decays to less than 8.0 Volts, with an output current of 60 milliamps and output ripple of less than 200 millivolts. The inductor value and timing should be adjusted for different applications (see the  $I_{OUT}$  equation). The maximum input voltage can be increased by adding a zener diode in series with the supply lead of each IC.

The oscillators are slaved together, using the low battery detection circuitry of the 4391, in order to reduce the electromagnetic interference that would otherwise be caused by having separate unsynchronized oscillator frequencies. The square wave output is available and can be used for other circuitry if the current drawn is low. Also available is the +1.25 Volt DC reference voltage (limit current to 100 $\mu$ A).

A circuit that allows higher input voltages and will deliver more power to the load than the standard application circuit is shown in (Figure 4). The circuit has an external switch transistor, Q1, with a higher maximum current rating than the internal switch transistor. Also, the optional zener diode D2 steps down the applied input voltage. If D2 is not used it should be shorted out.

The circuit operation is similar to that of the standard application circuit, except that the internal switch transistor now drives the base of the external transistor. Note that the on switch voltage will now equal the switch voltage of the internal transistor plus the  $V_{BE}$  of Q1. This higher switch voltage will reduce the circuits ability to supply output current with low input voltages. The external transistor circuit of (Figure 3) does not have this problem.

Note that even though the addition of the zener raises the maximum allowable supply voltage, it does not alter the maximum change of supply voltage, which equals the maximum supply voltage minus the minimum supply voltage ( $\Delta V = 30V - 3.0V = 27V$ ). So, with a 10V zener diode, the supply voltage range will be 13V to 40V.



**Figure 4. High Power/High Voltage Application Circuit**  
(Component Values Shown are for -5.0V Output)



# Precision High Speed Latching Comparator

4805

## Features

- 22nS Propagation Delay
- Low Offset Voltage —  $100\mu\text{V}$
- Low Offset Current —  $15\text{nA}$
- TTL Compatible Latch
- TTL Output

## Description

The RC/RM4805 is the ideal comparator for high speed, high precision applications. The input errors are factory trimmed to less than 1/10 LSB of a 12 bit, 10 volt system. The latch function allows the System Designer additional flexibility. When the latch input is a TTL low, the

comparator functions normally. When the input is raised to a TTL high; the comparator output is latched in its current state. The latch functions over the full military temperature range.

The 4805 is ideal for ultra precise, very fast system designs. Typical applications include successive approximation A/D converters of 12 or more bits, zero crossing detectors, high speed sampling, or window detectors.

The 4805 high speed comparator is functionally equivalent to the more popular comparators, HA-4950, AM686, SE527, CMP-05 and  $\mu\text{A760}$ . Propagation delay is 35nS with a 1/2 LSB overdrive in a 12 bit, 10 volt system.

## Connection Information

**TO-99  
Metal Can  
(Top View)**

65-00505A

**8-Lead  
Dual-in-Line  
Ceramic Package  
(Top View)**

65-00506A

Pin	Function
1	Ground
2	Non-Inverting Input
3	Inverting Input
4	V-
5	NC
6	Latch Enable
7	V <sub>OUT</sub>
8	V+

## Mask Pattern

Die Size: 51 x 67 mils  
 Min. Pad Dimensions: 4 x 4 mils

## Thermal Characteristics

	8-Lead Ceramic DIP	8-Lead TO-99 Metal Can
Max. Junction Temp.	175°C	175°C
Max. P <sub>D</sub> T <sub>A</sub> < 50°C	833mW	658mW
Therm. Res. $\theta_{JC}$	45°C/W	50°C/W
Therm. Res. $\theta_{JA}$	150°C/W	190°C/W
For T <sub>A</sub> > 50°C Derate at	8.33mW per °C	5.26mW per °C



## Typical Applications

Conversion Time			
	8-Bit	10-Bit	12-Bit
RM4805 Response	20nS	22nS	50nS
DAC Settling*	135nS	135nS	500nS
SAR Delay	50nS	50nS	50nS
Total-Cycle	205nS	207nS	600nS
Number Cycles + Reset	x9	x11	x13
<b>Total Conversion Time</b>	<b>1.8μS</b>	<b>2.3μS</b>	<b>7.8μS</b>

65-00540A

\*Response will be affected by DAC's output capacitance and equivalent resistance.

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**+10V Precision Voltage References**

**REF-01**

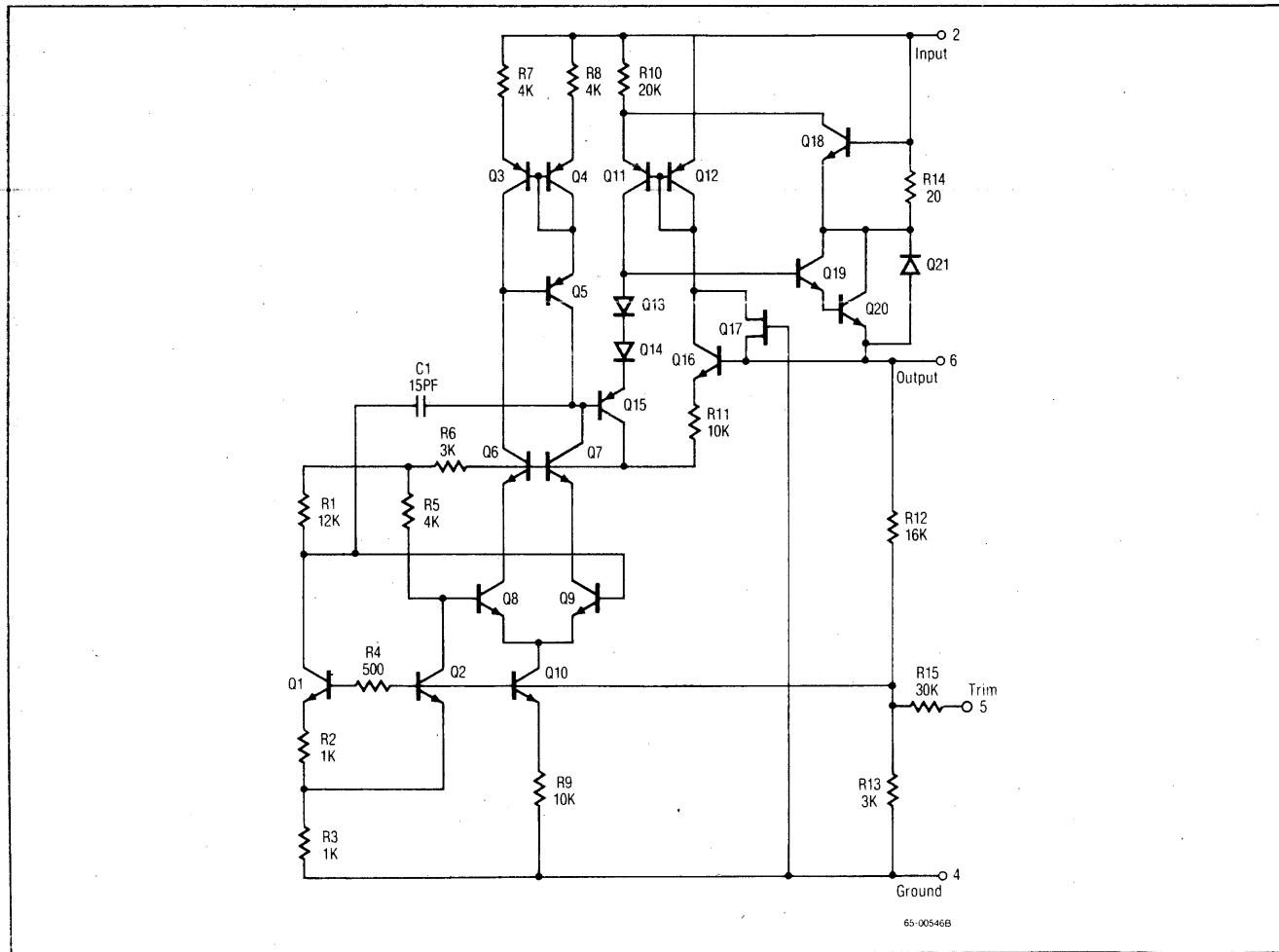
**Features**

- 10 Volt Output —  $\pm 0.3\%$
- Adjustable —  $\pm 3.0\%$
- Excellent Temperature Stability —  $3.0\text{ppm}/^\circ\text{C}$
- Low Noise —  $20\mu\text{V}_{\text{p-p}}$
- Wide Input Voltage Range — 12V to 40V
- No External Components
- Short Circuit Proof
- Low Power Consumption — 15mW

**Description**

The REF-01 Precision Voltage Reference contains a band-gap reference using thin-film resistors, a step-up amplifier, short circuit protection, and a zener-trim network. The REF-01's 10V output shows excellent stability for large changes of temperature, load current, and input voltage. An adjust pin is provided that can change the output voltage by at least 3.0% with little effect on temperature coefficient.

**Simplified Schematic Diagram**



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**+5.0V Precision Voltage References**

**REF-02**

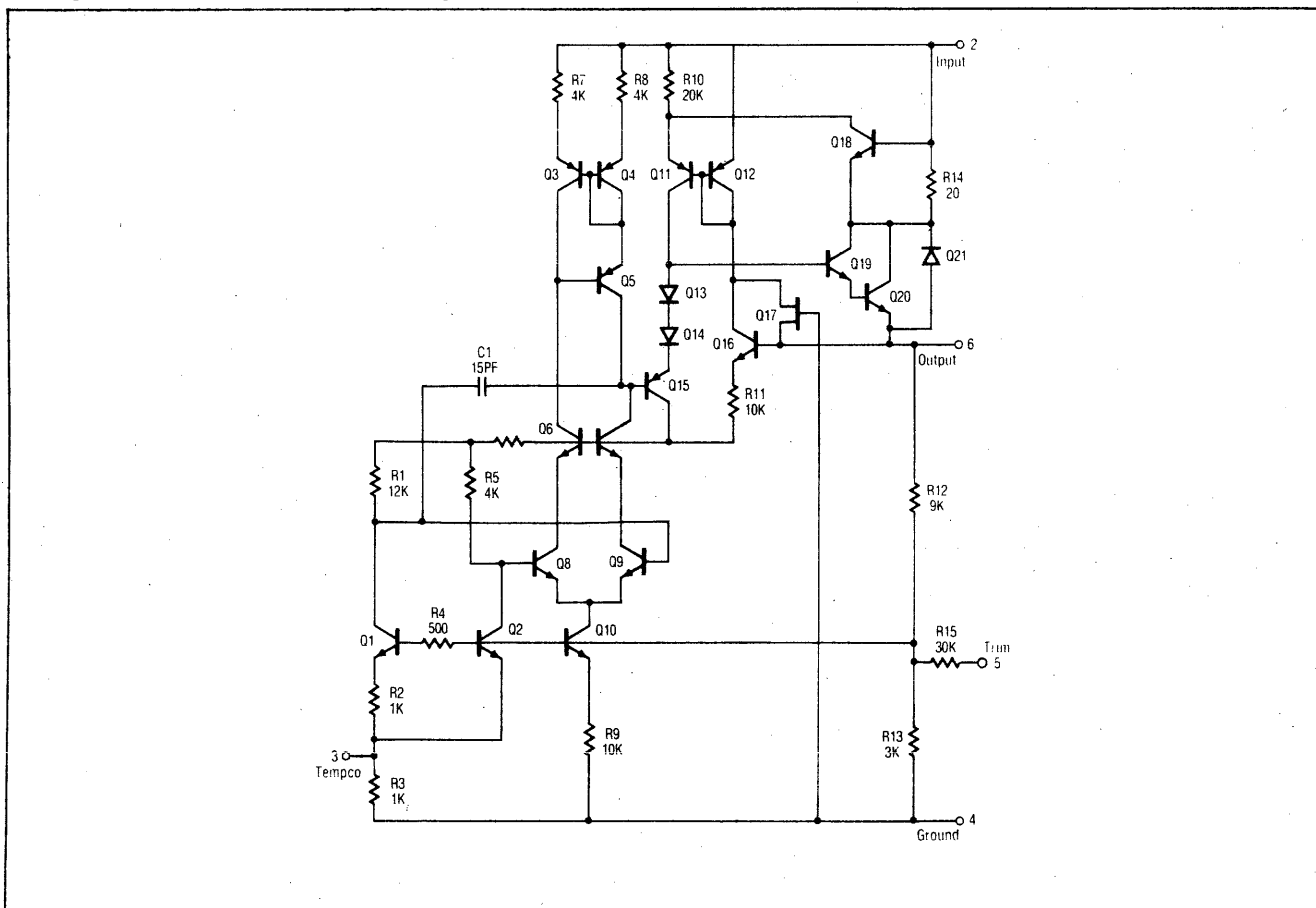
**Features**

- 5.0 Volt Output —  $\pm 0.3\%$
- Adjustable —  $\pm 3.0\%$
- Excellent Temperature Stability —  $3.0\text{ppm}/^\circ\text{C}$
- Low Noise —  $10\mu\text{V}_{\text{p-p}}$
- Wide Input Voltage Range — 7.0V to 40V
- No External Components
- Short Circuit Proof
- Low Power Consumption — 15mW

**Description**

The REF-02 Precision Voltage Reference contains a band-gap reference using thin-film resistors, a step-up amplifier, short circuit protection, and a zener-trim network. The REF-02's 5.0V output shows excellent stability for large changes of temperature, load current, and input voltage. An adjust pin is provided that can change the output voltage by at least 3% with little effect on temperature coefficient. A tempco pin also provides a voltage that varies linearly with temperature, typically from +470mV to +830mV over the military temperature range.

**Simplified Schematic Diagram**



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# +2.5V Precision Voltage References

REF-03

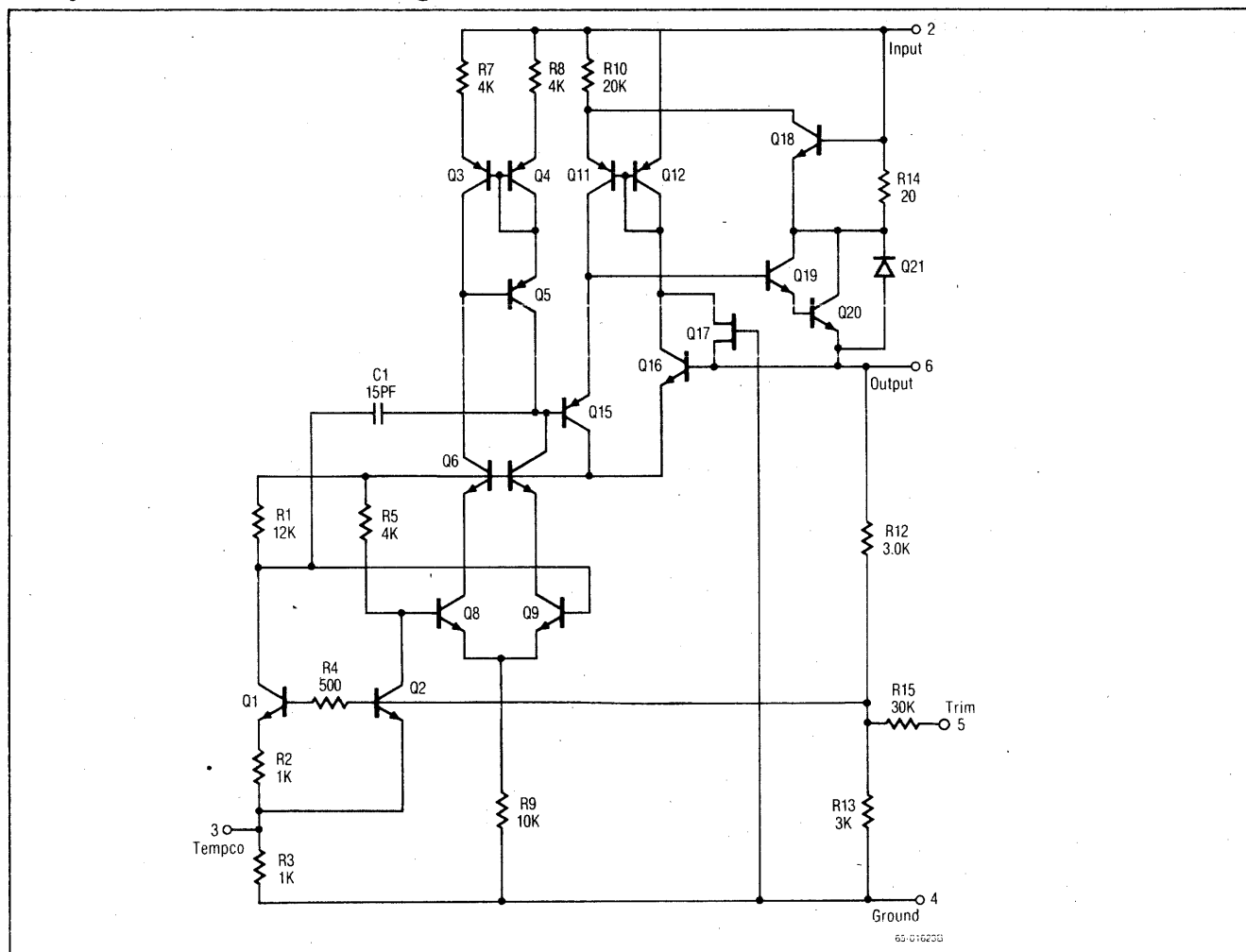
## Features

- 2.5 Volt Output —  $\pm 0.3\%$
- Adjustable —  $\pm 3\%$
- Excellent Temperature Stability —  $3.0\text{ppm}/^\circ\text{C}$
- Low Noise —  $5.0\mu\text{V}_{\text{p-p}}$
- Wide Input Voltage Range — 4.5V to 40V
- No External Components
- Short Circuit Proof
- Low Power Consumption — 15mW

## Description

The REF-03 Precision Voltage Reference contains a band-gap reference using thin-film resistors, a step-up amplifier, short circuit protection, and a zener-trim network. The REF-03's 2.5V output shows excellent stability for large changes of temperature, load current, and input voltage. An adjust pin is provided that can change the output voltage by at least 3.0% with little effect on temperature coefficient. A tempco pin also provides a voltage that varies linearly with temperature, typically from +470mV to +830mV over the military temperature range.

## Simplified Schematic Diagram



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# 10-Bit High Speed Multiplying D/A Converter

DAC-10

## Features

- Nonlinearity to 0.05% Max Over Temperature Range
- Low Full Scale Drift — 10ppm/°C
- Wide Range Multiplying Capability 1.0MHz bandwidth
- Wide Power Supply Range— +5.0V/-7.5V to ±18V
- Two Quadrant Multiplying
- High Output Compliance
- High Speed — 85nS

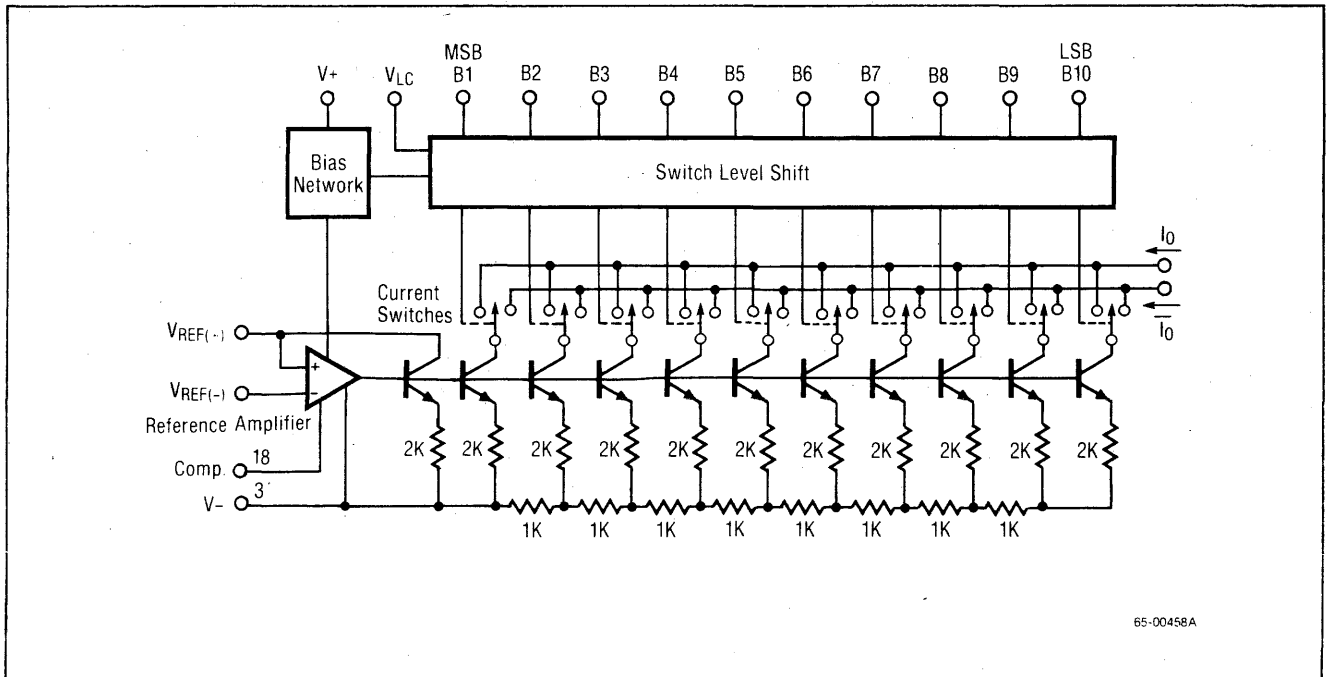
## Applications

- A/D Converters
- Servo Controls
- Waveform Generators
- Programmable Power Supplies
- High Speed Modems

## Description

The DAC-10 is a high speed, 10-bit, monolithic, multiplying Digital-to-Analog Converter. Settling times of 85nS are achieved with low power consumption and minimal output glitches. Full scale (10-bit) accuracy is achieved. The DAC-10 can be operated from almost any logic level input due to its adjustable ( $V_{LC}$ ) threshold. Monotonicity is guaranteed to 10 bits and nonlinearities of  $\pm 0.05\%$  are guaranteed over the full operating temperature range. Power consumption can be reduced to 85mW by lowering supply voltages to +5.0V to -7.5V. Operation at supply voltages up to  $\pm 18V$  does not appreciably affect device performance. Zener-Zap trimming is performed at wafer probe to optimize the converter's accuracy.

## Simplified Schematic Diagram



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# Micro-Power Switching Regulators

4191/2/3

## Features

- High Efficiency — 80% Typical
- Low Quiescent Current — 135 $\mu$ A
- Adjustable Output — 2.5V to 30V
- Output Current — 150mA
- Internal Reference — 1.31V
- Remote Shutdown Capabilities

## Description

Raytheon's micro-power switching regulators, 4191, 92 and 93, are the industry's first monolithic low power switching regulators available in a 8-lead mini-DIP, and designed specifically for battery operated instruments. They each contain a 1.3V temperature compensated band-gap reference, adjustable free running oscillator, voltage comparator, low battery detection circuitry and a 150mA switch transistor with all of the functions required to make a complete low power switching regulator.

These regulators can achieve up to 80% efficiency in most applications while being able to operate over a wide input supply voltage range, 2.4V to 24V, at an ultra low quiescent drain of 135 $\mu$ A.

These regulators have a free running oscillator which provides the drive circuitry for the on chip 150mA switch transistor. The 100Hz to 75kHz oscillator frequency is determined by an external capacitor on pin 2.

These universal regulators can be used as a building block in three basic applications: step-up, step-down, inverting.

In their most popular configuration, step-up, these regulators can use fewer battery cells, making available more board space and lengthen battery life.

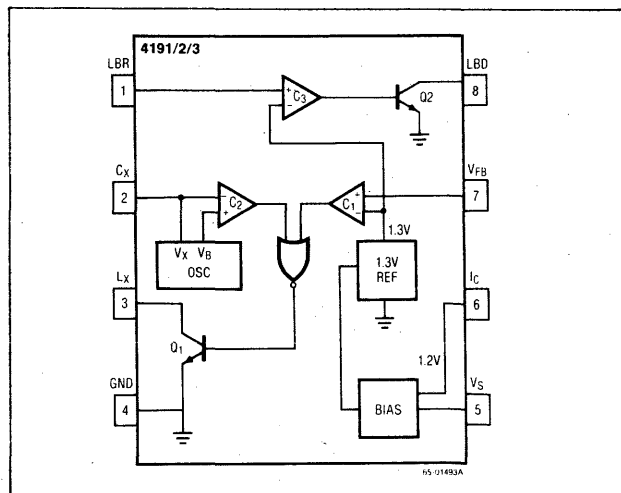
A practical example of these advantages would be an instrument designed to operate from a nominal 9 volt supply voltage. If this instrument were powered with just 7 components; a steering diode, an inductor, two resistors, a capacitor, a

4191, 92 or 93 and a 9 volt battery, it would receive a continuous 9 volts until the battery had decayed to a terminal voltage of 2.4 volts. If board space is at a premium, then the designer could remove the 9 volt source and replace it with a single 3 volt Ni-Cad battery without making any other adjustment(s) to the circuitry or affecting the overall operation of this instrument.

The open collector output transistor of these regulators provide the designer a method to have a "Low Battery Indicator" output current that can be used to signal a LCD whenever the battery voltage drops below a programmed level. This programmed level is determined by the designer with the selection of external resistors connected to pin 1.

The 4191, 92 and 93 offer the designer a mechanism to shut down this chip by connecting a zener diode in series with the I<sub>C</sub> lead (pin 6). This will shut down the chip whenever the battery voltage drops below the predetermined level. In this application, the circuit can be designed to signal a LCD whenever the battery voltage has dropped to the first programmed level, then shut itself off when the battery has dropped to the second level.

## Functional Block Diagram



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# Complete High-Speed 12-Bit Monolithic Digital-to-Analog Converter DAC-565

## Features

- Nonlinearity 1/2 LSB —0.012%
- Differential Nonlinearity 13 bits — 0.012%
- Settles to 1/2 LSB in 200nS
- On Chip Bandgap Reference
- Linearity Guaranteed Over Temperature
- Low Power: 225mW Including Reference
- Direct Interface to All Major Logic Families

## Description

The DAC-565 is a fast 12-bit digital-to-analog converter consisting of 12-bit precision, high-speed bipolar current steering switches, laser-trimmed thin film resistor network, stable bandgap reference, span and bipolar offset resistors to produce an accurate analog output current.

The high performance and flexibility of the DAC-565 was achieved through circuit design and layout, a thin film resistor process and interactive computer-controlled laser trimming. The DAC-565 settles to 1/2 LSB in 200nS, typical, with a maximum settling time of 500nS. Accuracy is specified at a maximum of 1/2 LSB for all grades.

This high speed and accuracy coupled with the inherent high output impedance make the DAC-565 the ideal DAC for high-speed display drivers, high speed control systems, and with the RC4805 high-speed latching comparator, analog-to-digital converters.

The bandgap reference is laser trimmed to optimize both temperature drift and absolute output voltage. The current sourcing capability of the DAC-565 reference (10mA typical) allows this reference to drive the DAC and also other peripheral circuit elements. Typical reference drift is better than 15ppm/°C (S and J grade).

The DAC-565 is available in 3 performance grades. The DAC-565J and DAC-565D grades are specified over 0°C to +70°C, while the DAC-565S grade is specified over the -55°C to +125°C temperature range.

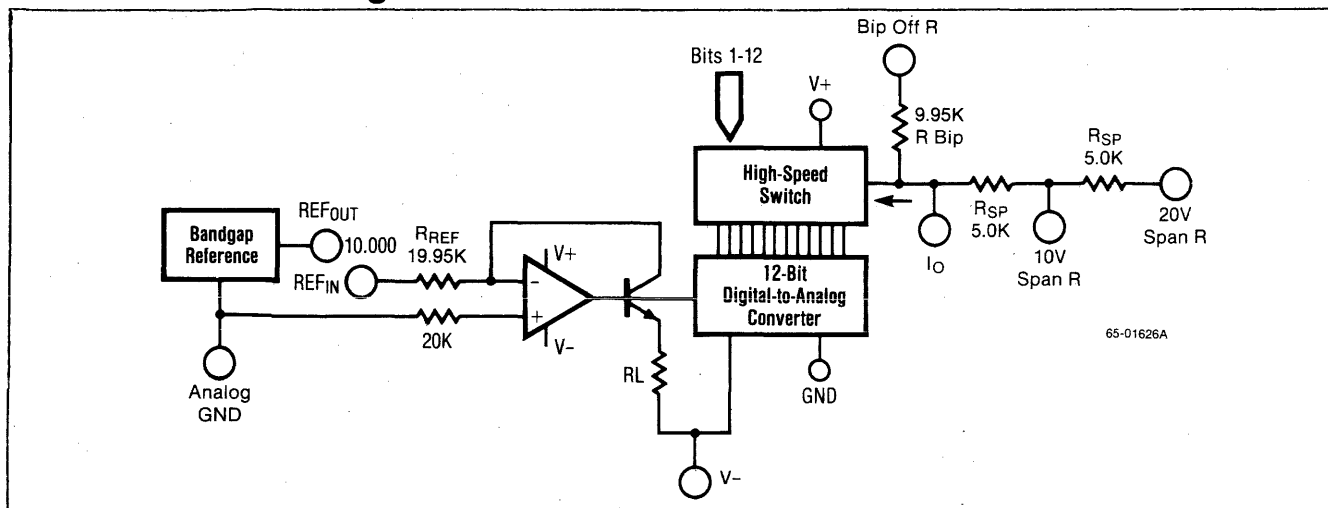
## Product Highlights

- The DAC-565 is a self-contained DAC and voltage reference with application resistors on a monolithic chip.
- The device incorporates interdigitizing which minimizes the effects of thin film sputtering, thermal, and diffusion gradients in the most critical portions of the design. Therefore, excellent linearity distributions are achieved prior to trimming, thus insuring optimal stability of the nonlinearity over temperature as well as stability versus time.
- The thin film resistors have a trim tab which is distant from the main body of the resistor. This resistor geometry insures near perfect nonlinearities after trim as well as minimum damage due to laser trimming.
- The internal reference is laser trimmed to 10 Volts with a ±1.0% maximum error. The reference voltage is available externally and can provide typically 10mA beyond that required for the reference and bipolar offset resistors.

The DAC-565 contains SiCr thin film application resistors which can be used with either an external op amp to provide a precision voltage output DAC or as input resistors for a successive approximation A/D converter. The resistors are inherently matched and are laser trimmed to guarantee minimum full scale and bipolar offset errors.

- The DAC-565S grade guarantees linearity and monotonicity over the -55°C to +125°C range and is available fully processed to MIL-STD-883, Level B.

## Functional Block Diagram



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# 12-Bit High Speed Multiplying D/A Converter

DAC-6012

## Features

- Differential Nonlinearity — 0.012% (13 Bits)
- Guaranteed Monotonicity to 12 Bits Over Temperature
- Relative Accuracy — 0.05% All Grades
- Fast Settling Time — 250nS to  $\pm 0.5$  LSB
- Full Scale Output Current — 4.0mA
- Complementary Current Outputs
- Output Compliance — -5.0V to +10V
- Full Scale Tempco —  $\pm 10$ ppm/ $^{\circ}$ C
- Power Consumption — 230mW
- Direct Interface to All Major Logic Families
- Standard Processing without Resistor Trimming
- Low Cost

## Description

The Raytheon DAC-6012 series of monolithic Multiplying Digital-to-Analog Converters guarantee differential nonlinearity to better than  $\pm 0.5$  LSB (0.012%) for the 6012A and  $\pm 1.0$  LSB (0.025%) for the 6012 over the full military and commercial temperature ranges. In addition to the excellent Differential Nonlinearity specifications, the 6012 series also include many features that previously were found in expensive hybrid modules or required full use of monolithic thin film laser or zener zap trimming techniques.

The Raytheon DAC-6012 incorporates a segmented design technique which reduces the requirement for high accuracy resistor ladder networks as an integral part of the DAC. The DAC-6012 design is structured with a 3-bit

segment decoder, 5-bit master R-2R ladder DAC and 4-bit Slave DAC. This circuit configuration actually contains less ladder resistors than the traditional R-2R ladder approach as well as effectively improving the accuracy of the ladder resistors by a factor of 8.

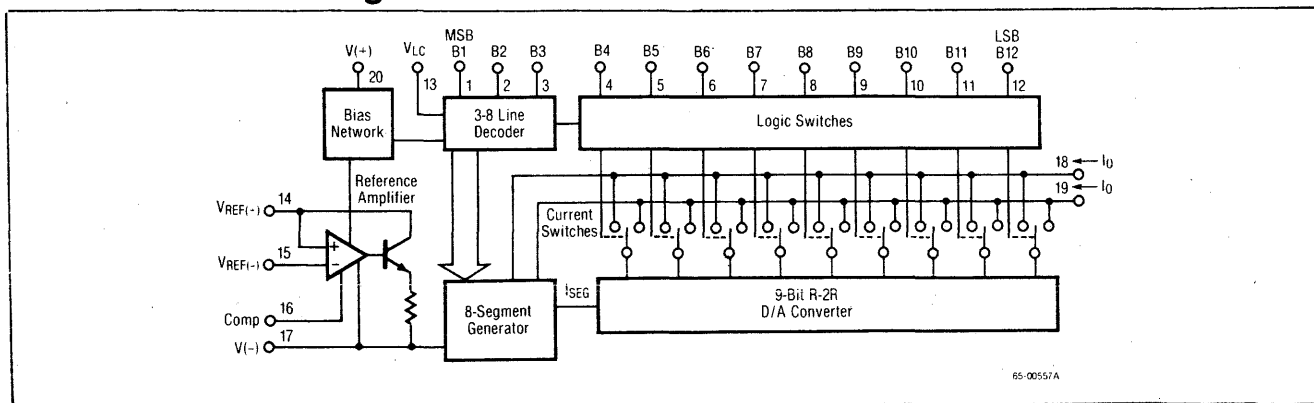
The performance of the DAC-6012 is virtually independent of supply voltage variations due to the inherent nature of its design and processing. As an example, the DAC-6012 may be operated at any voltage from +4/-10V to  $\pm 18$ V with minimal effect on the full scale current, DNL, relative accuracy and settling time. The 5M $\Omega$  output impedance and -5.0V to +10V compliance range make the DAC-6012 ideal for high speed applications where output load resistors can be used in place of an output interface amplifier.

The complementary current outputs of the DAC-6012 are useful in symmetrical offset DAC applications and A/D converters requiring constant current loads to ensure significant reduction of switching transients.

In conjunction with the REF-01 and REF-02 voltage references, and the 4805 fast precision voltage comparator, the DAC-6012 can be used as the main building block in a wide variety of data conversion applications.

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## Functional Block Diagram



Raytheon LINEAR





# 8-Bit High Speed Multiplying D/A Converter

DAC-08

## Features

- Fast Settling Output Current — 85nS
- Full Scale Current Prematched to  $\pm 1.0$  LSB
- Direct Interface to TTL, CMOS, ECL, HTL, PMOS
- Nonlinearity to  $\pm 0.1\%$  Max. Over Temperature Range
- High Output Impedance and Compliance —  $-10V$  to  $+18V$
- Differential Current Outputs
- Wide Range Multiplying Capability — 1.0MHz Bandwidth
- Low FS Current Drift —  $\pm 10$ ppm/ $^{\circ}C$
- Wide Power Supply Range —  $\pm 4.5V$  to  $\pm 18V$
- Low Power Consumption — 33mW @  $\pm 5.0V$
- Low Cost

## Description

The DAC-08 series of 8-bit monolithic multiplying Digital-to-Analog Converters provide very high speed performance coupled with low cost and outstanding applications flexibility.

Advanced circuit design achieves 85nS settling times with very low "glitch" and at low power consumption. Monotonic multiplying performance is attained over a wide 40 to 1 reference current range. Matching to within 1 LSB between reference and full scale currents eliminates the need for full scale trimming in most applications.

Direct interface to all popular logic families with full noise immunity is provided by the high swing, adjustable threshold logic inputs.

High voltage compliance dual complementary current outputs are provided, increasing versatility and enabling differential operation to effectively double the peak-to-peak output swing. In many applications, the outputs can be directly converted to voltage without the need for an external op amp.

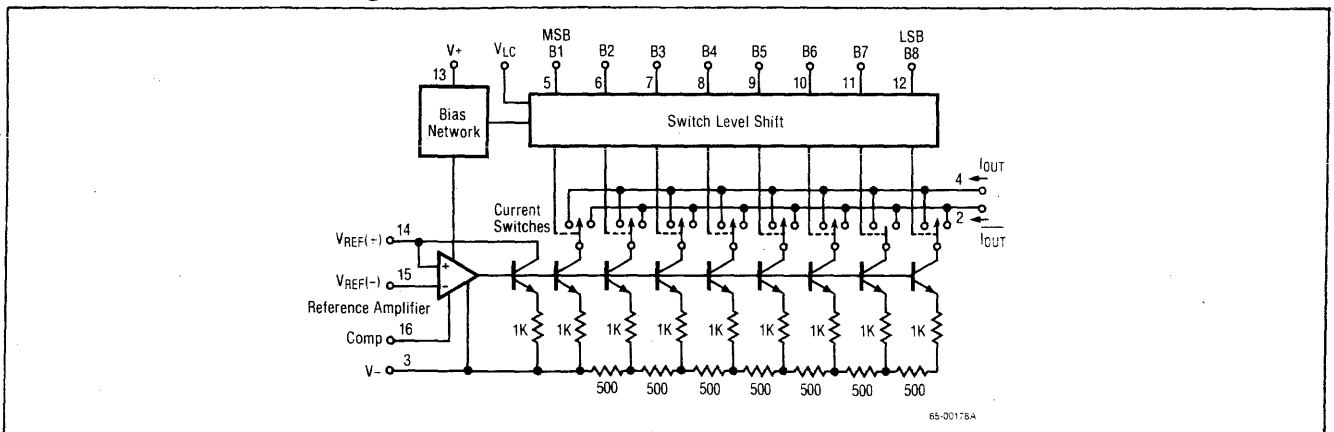
All DAC-08 series models guarantee full 8-bit monotonicity, and nonlinearities as tight as  $\pm 0.1\%$  over the entire operating temperature range are available. Device performance is essentially unchanged over the  $\pm 4.5V$  to  $\pm 18V$  power supply range, with 33mW power consumption attainable at  $\pm 5.0V$  supplies.

The compact size and low power consumption make the DAC-08 attractive for portable and military/aerospace applications; devices processed to MIL-STD-883A, Level B are available.

DAC-08 applications include 8-bit, 1.0 $\mu$ S A/D converters, servo-motor and pen drivers, waveform generators, audio encoders and attenuators, analog meter drivers, programmable power supplies, CRT display drivers, high speed modems and other applications where low cost, high speed and complete input/output versatility are required.

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## Functional Block Diagram



High Performance Operational Amplifiers

Device	Description	Maximum Input Specifications			Typical Input Noise Specifications				Typ. Slew Rate (V/ $\mu$ S)	Typ. Unity Gain BW (MHz)	Max. Sply Volt. ( $\pm$ V)	Pins	Temp Range
		Offset Volt. (mV)	Offset Cur. (nA)	Bias Cur. (nA)	Volt. Density nV/ $\sqrt$ Hz	Low Freq. Corner en (Hz)	Cur. Density PA/ $\sqrt$ Hz	Low Freq. Corner in (Hz)					
OP-05	Single Instrumentation Grade	0.500	2.8	$\pm$ 3.0	10.0	10	0.14	50	0.17	0.5	22	8	M, C
OP-05A	Single Instrumentation Grade	0.150	2.0	$\pm$ 2.0	10.0	10	0.14	50	0.17	0.5	22	8	M, C
OP-05C	Single Instrumentation Grade	1.300	6.0	$\pm$ 7.0	10.2	10	0.15	50	0.17	0.5	22	8	M, C
OP-05E	Single Instrumentation Grade	0.500	3.8	$\pm$ 4.0	10.0	10	0.14	50	0.17	0.5	22	8	M, C
OP-07	Single Ultra Low Offset Voltage	0.075	2.8	$\pm$ 3.0	10.0	10	0.14	50	0.17	0.6	22	8	M
OP-07A	Single Ultra Low Offset Voltage	0.025	2.0	$\pm$ 2.0	10.0	10	0.14	50	0.17	0.6	22	8	M
OP-07C	Single Ultra Low Offset Voltage	0.150	6.0	$\pm$ 7.0	10.2	10	0.15	50	0.17	0.6	22	8	C
OP-07D	Single Ultra Low Offset Voltage	0.150	6.0	$\pm$ 12	10.3	10	0.15	50	0.17	0.6	22	8	C
OP-07E	Single Ultra Low Offset Voltage	0.075	3.8	$\pm$ 4.0	10.0	10	0.14	50	0.17	0.6	22	8	C
OP-27A	Single Ultra Low Noise	0.025	35	$\pm$ 40	3.0	3.0	0.4	140	2.8	8.0	22	8	M, C
OP-27B	Single Ultra Low Noise	0.060	50	$\pm$ 55	3.0	3.0	0.4	140	2.8	8.0	22	8	M, C
OP-27C	Single Ultra Low Noise	0.100	75	$\pm$ 80	3.2	3.0	0.4	140	2.8	8.0	22	8	M, C
OP-27E	Single Ultra Low Noise	0.025	35	$\pm$ 40	3.0	3.0	0.4	140	2.8	8.0	22	8	M, C
OP-27F	Single Ultra Low Noise	0.060	50	$\pm$ 55	3.0	3.0	0.4	140	2.8	8.0	22	8	M, C
OP-27G	Single Ultra Low Noise	0.100	75	$\pm$ 80	3.2	3.0	0.4	140	2.8	8.0	22	8	M, C
OP-37A*	Single High Slew Rate Low Noise	0.025	35	$\pm$ 40	3.0	3.0	0.4	140	11	63	22	8	M, C
OP-37B*	Single High Slew Rate Low Noise	0.060	50	$\pm$ 55	3.0	3.0	0.4	140	11	63	22	8	M, C
OP-37C*	Single High Slew Rate Low Noise	0.100	75	$\pm$ 80	3.2	3.0	0.4	140	11	63	22	8	M, C
OP-37E*	Single High Slew Rate Low Noise	0.025	35	$\pm$ 40	3.0	3.0	0.4	140	11	63	22	8	M, C
OP-37F*	Single High Slew Rate Low Noise	0.060	50	$\pm$ 55	3.0	3.0	0.4	140	11	63	22	8	M, C
OP-37G*	Single High Slew Rate Low Noise	0.100	75	$\pm$ 80	3.2	3.0	0.4	140	11	63	22	8	M, C
RC714	Single Precision	0.075	2.8	$\pm$ 3.0	9.6	3.0	0.12	140	0.17	0.5	22	8	C
RC714C	Single Precision	0.150	6.0	$\pm$ 7.0	9.8	3.0	0.13	140	0.17	0.5	22	8	C
RC714E	Single Precision	0.075	3.8	$\pm$ 4.0	9.6	3.0	0.12	140	0.17	0.5	22	8	C
RC714L	Single Precision	0.250	20	$\pm$ 30	9.8	3.0	0.13	140	0.17	0.5	22	8	C
RC3078	Single Micropower	4.5	32	170	19.0	100	1.0	200	1.5	0.1	15	8	M, C
RC3078A	Single Micropower	3.5	2.5	12	36.0	100	0.4	200	1.5	0.1	15	8	M, C
RC5534	Single High Performance Low Noise	4.0	300	1500	4.0	100	0.6	200	13.0	10.0	22	8	M, C
RC5534A	Single High Performance Low Noise	4.0	300	1500	3.5	100	0.4	200	13.0	10.0	22	8	M, C
RC2041	Dual High Performance Low Noise	3.0	200	500	5.0	20	0.4	200	3.0	7.0(4.0)	18	8	C
RC2043	Dual High Performance Low Noise	3.0	200	1000	5.0	20	0.4	200	6.0	14(8.0)	18	8	C
RC4560	Dual High Performance	6.0	200	500	10.0	20	0.5	200	4.0	10(7.0)	18	8	C
RC4562	Dual High Performance	6.0	200	500	6.0	20	0.2	200	7.0	15(8.0)	18	8	C
RC4558	Dual High Gain	6.0	200	500	10.0	20	0.5	200	1.0	2.5	18	8	M, C
RC4559	Dual High Performance	6.0	100	250	10.0	20	0.18	200	2(1.5)	4.0(3.0)	18	8	M, C
RC4739	Dual Low Noise	6.0	200	500	10.0	20	0.5	400	1.0	4.0	18	14	M, C
RC5532	Dual High Performance Low Noise	2.0	150	800	5.0	100	0.7	200	8.0	10.0	22	8	M, C
RC5532A	Dual High Performance Low Noise	2.0	150	800	5.0	100	0.7	200	8.0	10.0	22	8	M, C
RC4556	Dual High Performance	6.0	200	500	10.0	20	0.5	200	3.0	8.0(5.0)	18	8	C

Notes:

- ( ) denotes guaranteed specifications
- M = Military/883B -55°C to +125°C
- C = Commercial 0°C to +70°C
- = Denotes preliminary data

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LINEAR

Raytheon



## Linear Integrated Circuits

The RCA line of linear integrated circuits (LIC's) covers a broad range of solid-state devices, including BiMOS, bipolar, CMOS, and MOS/FET types for a wide variety of applications in industrial, military, and consumer equipment. These circuits are available in a wide variety of package styles, including hermetic TO-5; dual-in-line plastic, ceramic, and frit-seal ceramic; and TO-220-style (Versa-V) plastic. RCA also offers high-reliability versions of a broad range of these standard circuits processed in accordance with MIL-STD-883 and MIL-M-38510.

### BiMOS Circuits

RCA BiMOS op-amps and comparators utilizing both bipolar and p-channel MOS transistors on a common monolithic substrate establish new performance standards in virtually all op-amp categories: general-purpose, FET input, wideband (4 to 70 MHz), micropower (down to 1.5  $\mu$ W), high-current (to 22 mA), and in comparator applications, particularly in those requiring the added convenience and flexibility of two single-supply comparators in one package with operation up to 44 volts.

### Operational Amplifiers

RCA offers a broad line of operational amplifiers, including general-purpose types in single, dual, and quad configurations; types including mixed technologies (e.g., CMOS, PMOS and bipolar, called BiMOS); variable types often referred to as programmable op-amps; micropower; wide-band; high-voltage; high-current; and precision types.

### Voltage Comparators

RCA's line of BiMOS and bipolar comparators includes single, dual, quad, and programmable types for a wide variety of applications in instrumentation, communications, and industrial equipment.

### Data Conversion Circuits

This category includes A/D converters, 7-segment decoder/drivers, and display drivers.

### Arrays

RCA's broad line of arrays includes transistor types (bipolar and CMOS), transistor amplifier types, transistor/diode types, and diode types in a wide variety of configurations: n-p-n, p-n-p, n-p-n/p-n-p, Darlington, superbeta, programmable unijunction transistor (PUT) and SCR.

### Power Control Circuits

This category includes an automotive ignition switch, power amplifiers, programmable Schmitt triggers, solenoid and motor drivers, universal controller, voltage regulators and zero-voltage switches.

### Differential Amplifiers

The RCA line of differential amplifiers includes a broad variety of general-purpose devices in which the basic signal-processing function is accomplished by a single differential-amplifier stage.

### Special Function Circuits

These types include automotive circuits, broadband (video) amplifiers, four-quadrant multipliers, prescalers, single-chip detector alarm systems and a timer circuit.

### TV/CATV Circuits

These circuits include AFT, chroma systems, horizontal and vertical deflection systems, sync/AGC circuits, luminance processors, multiplex decoders, PIX IF, remote control, sound IF, tuning and Videodisc circuits.

**For a summary of available products refer to the Linear Integrated Circuits product guide CDL-820J for linear integrated circuits and to MOS Field-Effect Transistors catalog MOS-160D for MOS devices. For complete technical data refer to Linear Integrated Circuits DATABOOK SSD-240B.**

## LIC Operational Amplifiers

### BiMOS Types

- **CA3193 BiMOS Precision Operational Amplifier**
- **CA3493 BiMOS Precision Operational Amplifier**

The CA3193 series and CA3493 series are ultra-stable, precision-instrumentation operational amplifiers that employ both PMOS and bipolar transistors on a single chip. They are internally phase-compensated and provide a gain-bandwidth product of 1.2 MHz. The CA3193 series are pin-compatible with the industry 741 series and many other IC op amps, and may be used as replacements for 741-series in most applications.

The CA3493 series are pin-compatible with many industry types such as 725, 108A, OP-5, OP-7, and LM714 where positive nulling is employed.

#### Features

- Low  $V_{IO}$ :
  - 75  $\mu$ V max. (CA3493B)
  - 200  $\mu$ V max. (CA3493A)
  - 500  $\mu$ V max. (CA3493)
- Low  $\Delta V_{IO}/\Delta T$ :
  - 2  $\mu$ V/ $^{\circ}$ C max. (CA3493B)
  - 3  $\mu$ V/ $^{\circ}$ C max. (CA3493A)
  - 5  $\mu$ V/ $^{\circ}$ C max. (CA3493)
- Low  $I_{IO}$  and  $I_i$

- **CA3420 Low Supply Voltage, Low-Input Current BiMOS OP-AMP**
- **CA3440 Nanopower BiMOS OP AMP**

The CA3420, CA3420A and CA3420B BiMOS operational amplifiers feature gate-protected PMOS transistors in the input circuit to provide very high input impedance, very low input currents (less than 1 pA). The internal bootstrapping network features a unique guardbanding technique for reducing the doubling of leakage current for every 10 $^{\circ}$  C increase in temperature. The CA3420 series operates at total supply voltages from 2 to 20 volts either single or dual supply.

#### Features

- 2 V Supply at 300  $\mu$ A Supply Current
- 0.04 pA (typ) Input Current (Essentially Constant to 85 $^{\circ}$ C)
- Rail-to-Rail Output Swing (Drive  $\pm$ 2 mA into 1 K $\Omega$  Load)
- Pin Compatible with 741 Op-Amp
- Low Cost 8-Lead Minidip, TO-5
- Output driving current 1.5 ma. min. (provided by non-linear current mirrors)

#### Applications

- pH Probe Amplifiers
- Picoammeters
- Electrometer (High Z) instruments
- Portable Equipment
- Inaccessible Field Equipment
- Battery Dependent Equipment (Medical and Military)
- Function generators
- Oscillators

- Low  $\Delta I_{IO}/\Delta T$ :
  - 50 pA/ $^{\circ}$ C max. (CA3493B)
  - 150 pA/ $^{\circ}$ C max. (CA3493)
- Low  $\Delta I_i/\Delta T$ :
  - 0.5 nA/ $^{\circ}$ C max. (CA3493B)
  - 3.7 nA/ $^{\circ}$ C max. (CA3493)
- Extremely high gain:
  - 120 dB min. (CA3493B)
  - 110 dB min. (CA3493A)
  - 100 dB min. (CA3493)
- Low  $V_{IO}$  vs. time
- High CMRR and PSRR
- Internally compensated: 1.2-MHz gain-bandwidth product
- Low input noise: 0.1 Hz to 10 Hz
  - Noise voltage: 0.36  $\mu$ V<sub>p-p</sub> typ.
  - Noise current: 12 pA<sub>p-p</sub> typ.

#### Applications

- Thermocouple preamplifiers
- Strain-gauge bridge amplifiers
- Summing amplifiers
- Differential amplifiers
- Bilateral current sources
- Log amplifiers
- Differential voltmeters
- Precision voltage references
- Active filters
- Buffers
- Integrators
- Sample-and-hold circuits
- Low frequency filters

The CA3440-series BiMOS OP AMP features gate-protected PMOS transistors in the input circuit to provide very high input impedance and very low input current (10 pA). These devices operate at total supply voltages from 5 to 15 volts and can be operated over the temperature range from -55 C to +125 C. Their virtues are programmability and very low standby power (300 nW).

#### Features

- 300-nW (typ.) standby power at  $V^+ = 5$ V
- Supply current, BW, slew rate programmable using external resistor
- 10-pA (typ.) input current
- 5.0 to 15-V supply ( $\pm$ 2.5 to  $\pm$ 7.5V)/5-12 V single supply operation
- Output drives typical bipolar-type loads
- Low-cost 8-lead Mini-DIP, TO-5
- High  $I_{out}/I_{standby}$  ratio at 900nW standby (can on demand sink 3.1 mA max (2 K load), source 3mA)
- Very low total standby currents (50 nA)
- Pin-compatible with 741 op amp

#### Recommended Applications

- Micropower Band-Gap Reference supply circuits
- Micropower Amplifier circuits
- Multimeters
- Audio amplifiers
- Solar-cell powered systems
- Voice-squelched amplifiers
- Battery and portable instruments
- Smoke detectors and security systems
- Biomedical comparators — buffers
- Active filters
- Multivibrators

 LINEAR  
RCA

# LIC Operational Amplifiers

## BiMOS Types

### • CA3260 Dual OP AMP

The RCA CA3260 series are operational amplifiers that combine the advantages of both CMOS and bi-polar transistors on a monolithic chip. The CA3260 series are dual versions of the CA3160 series.

#### Features

- Dual version of the CA3160
- MOS/FET input stage provides:
  - very high  $Z_i = 1.5 T\Omega$  ( $1.5 \times 10^{12}\Omega$ ) typ.
  - very low  $I_i = 5$  pA typ. at 15-V operation
  - $= 2$  pA typ. at 5-V operation
- Common-mode input-voltage range includes negative supply rail; input terminals can be swung 0.5 V below negative supply rail
- CMOS output stage permits signal swing to either (or both) supply rails
- Low  $V_{io}$ : 2 mV max. (CA3260B)
- Wide BW: 4 MHz typ. (unity-gain crossover)

High SR: 10 V/ $\mu$ s typ. (unity-gain follower)

High output current ( $I_o$ ): 20 mA typ.

- High  $A_{OL}$ : 320,000 (110 dB) typ.
- Internal phase compensation for unity gain
- Same pin-out as CA358, CA1458, CA1558

#### Applications

- Ground-referenced single-supply amplifiers
- Fast sample-hold amplifiers
- Long-duration timers/monostables
- Ideal interface with digital CMOS
- High-input-impedance wideband amplifiers
- Voltage followers (e.g., follower for single-supply D/A converter)
- Voltage regulators (permits control of output voltage down to zero volts)
- Wien-Bridge oscillators
- Voltage-controlled oscillators
- Photo-diode sensor amplifiers

### • CA3240 Dual-Unit Extended-Voltage-Range Op-Amp

### • CA3140 Extended-Voltage-Range Op-Amp

- very high input impedance: 1.5 T $\Omega$  typ.
- very low input current: 10 pA typ. at  $\pm 15$  V
- low input offset voltage: as low as 2 mV max.
- wide common-mode input-voltage range: can be swung 0.5 V below negative rail
- output swing complements input common-mode range, permitting full utilization of low supply voltages (to 4 V) or to within 0.2 V when driving power transistors (no need for level-shifting circuits)
- PMOS input devices are protected by rugged bipolar diodes plus bipolar speed and high supply voltage.
- Operating capability: 4-44 V, single or dual supply

#### Features

- Internally compensated; needs no external components
- Wide bandwidth (4.5-MHz unity gain) makes possible low-cost video and audio circuits
- Fast settling time (1.4  $\mu$ s typ. to 10 mV) for low-cost sample-and-hold and data-acquisition use
- Directly replaces industry types 741 (CA3140) and 1458 (CA3240) in most applications
- Characterized for  $\pm 15$ -V operation and for TTL supply systems down to 4 V
- High voltage-follower slew rate (9 V/ $\mu$ s)
- Strobable output stage
- Compared to BiFET devices, the CA3140 and CA3240 offer lower input currents, higher input resistance, improved offset current, and comparable offset voltage

#### COMPARISON CHART

Sym- bol ( $T_A =$ 25°C)	Characteristics and Limits						Units
	CA3240			1458			
	Min.	Typ.	Max.	Min.	Typ.	Max.	
$R_i$	30K	1500K	—	0.3	2	—	M $\Omega$
$I_i$	—	10	50	—	80K	500K	pA
$I_{io}$	—	0.5	30	—	20K	200K	pA
$V_{io}$	—	5	15	—	2	6	mV
SR	—	9	—	—	0.5	—	V/ $\mu$ s
$f_T$	—	4.5	—	—	1	—	MHz
$V_{ICR}$	-15	-15.5 to +12.5	+11	-12	$\pm 13$	+12	V
$V_{o\ p-p}$	-14	-14.4 to +13	+12	-10	$\pm 13$	+10	V
$A_{OL}$	20K	—	—	20K	—	—	V/V

#### Recommended Applications

- Ground-referenced single-supply amplifiers in auto and portable instrumentation
- Long-duration timers/multivibrators (micro-seconds, minutes, hours)
- Interface in 5-V TTL systems and other low-supply-voltage systems
- All standard operational-amplifier applications
- Peak detectors
- Comparators
- Active filters
- Portable supplies

# LIC Operational Amplifiers

## BiMOS Types

- CA3100 Wideband Op-Amp
- CA3160 Universal Op-Amp with Internal Compensation;
- CA3130 without Internal Compensation

CA3160 Three technologies on a single chip; CA3100 MOS/FET input stage, bipolar stage, and CMOS output stage

- CMOS output stage can swing output voltage to within 10 mV of either supply rail at very high values of load impedance

CA3130 PMOS and bipolar on a single chip

- Will handle both small and large signal inputs and still produce a true output virtually free of ringing and distortion

### Features of CA3130, CA3160

- MOS/FET input features (See CA3140, CA3240)
- Wide bandwidth, typically 15 MHz for CA3130, 4 MHz for CA3160
- High slew rate: 10 V/μs typ. (unity-gain follower)
- High output current (I<sub>o</sub>): 20 mA typ.
- High A<sub>OL</sub>: 320,000 (110 dB) typ.

### Features of CA3100

- High open-loop gain: 42 dB typ. at 1 MHz
- High unity-gain f<sub>T</sub>: 38 MHz typ.
- Wide power bandwidth (V<sub>o</sub>): 18 V<sub>p-p</sub> at 1.2 MHz
- High slew rate: 70 V/μs typ. (20-dB amplifier), 25 V/μs (unity-gain amplifier)
- Fast settling time: 0.6 μs typ.
- LM118, LM748, LM101 compatibility

### Recommended Applications

CA3130, CA3160

- High-input impedance comparators
- Ideal interface with digital CMOS
- Voltage followers
- Voltage regulators
- For other applications, see CA3140, CA3240

CA3100

- Video amplifiers
- Fast peak detectors
- Video pre-drivers
- Multivibrators
- Oscillators
- Meter driver amplifiers
- Oscillators

Symbol (TA = 25°C)	Characteristics and Limits						Units
	CA3100			CA3130/CA3160			
	Min.	Typ.	Max.	Min.	Typ.	Max.	
R <sub>i</sub>	—	—	—	30K	1500K	—	MΩ
I <sub>i</sub>	—	0.7†	2†	—	5	50	pA
I <sub>io</sub>	—	0.05†	0.4†	—	0.5	30	pA
V <sub>io</sub>	—	1	5	—	6	15	mV
SR	50	70	—	—	10	—	V/μs
f <sub>T</sub>	—	38	—	—	4	—	MHz
V <sub>icr</sub>	±12	+14 to -13	—	0	-0.5 to +12	+10	V
V <sub>o p-p</sub>	-9	+11 to -11	+9	-0.01	-0.002 to +13.3	+12	V
A <sub>OL</sub>	630	1.1K	—	—	50K	—	V/V

†microamperes

## CA080, CA081, CA082, CA083, CA084 Series

The CA080 series offer lower input bias and offset currents, higher input impedances, and wider bandwidth than their BiFET counterparts and other pin-compatible op amps.

### Features

- Very low input bias current (50 pA max.) and offset current (30 pA max.)
- Input impedance typically 10<sup>12</sup>Ω
- Low input offset voltage (2mV typ.)
- Wide common-mode input voltage range
- Lower power consumption: 42mW (single amp. channel)
- High slew rate: 13v/us typ.
- Unity-gain bandwidth = 5 MHz (typ.)
- Wide output voltage swing: 25 volts
- Low distortion
- Short-circuit protection
- Noise voltage (1 kHz): 40 nV/√Hz
- Direct replacement for industry type TL080 series in most applications

### Applications

- Inverters
- High-Q notch filters

## BiMOS Replacements for Industry TL080 Series BiFET Types

- IC preamplifiers
- Unity-gain absolute value amplifiers
- Sample and hold amplifiers
- Active filters
- Low-current amplification circuits
- Data acquisition equipment
- AC voltmeters (front end)

Char-acter-istics T <sub>A</sub> =25°C	CA080 CA081 CA082 CA083 CA084	CA080A CA081A CA082A CA083A CA084A	CA080B CA081B CA082B CA083B CA084B	CA080 CA081 CA082	CA080A CA081A CA082A	Units
T <sub>A</sub> Range	0-70	0-70	0-70	-55 to 125 °C		
V <sub>io</sub> (max.)	15	6	3	6	3	V
I <sub>i</sub> (max.)	50	40	30	40	40	pA
I <sub>io</sub> (max.)	30	20	10	20	20	pA
I <sup>+</sup> (max.)	2.8*	2.8*	2.8*	2.8*	2.8*	mA
V <sub>icr</sub> (min.)	±10	±12	±12	±12	±12	V
A <sub>OL</sub> (min.)	25K	50K	50K	50K	50K	V/V
f <sub>T</sub> (typ.)	5	5	5	5	5	MHz
SR (typ.)	13	13	13	13	13	V/μs

\*Per ampl CA3082, CA3083, CA3084

RCA LINEAR

# LIC Operational Amplifiers-OTAs

## CA3280 Dual Variable OP AMP

The CA3280 series consists of two variable operational amplifiers that are designed to substantially reduce the initial input offset voltage and the offset-voltage variation with respect to changes in programming current. This design results in reduced "AGC thump", an objectional characteristic of many AGC systems. Interdigitation, or cross coupling of critical portions of the circuit reduces the amplifier dependence upon thermal and processing variables.

The CA3280 has all the generic characteristics of an operational voltage amplifier except that the forward transfer characteristic is best described by transconductance rather than voltage gain, and the output is current, not voltage. The magnitude of the output current is equal to the product of transconductance and the input voltage. This type of operational transconductance amplifier has gained wide acceptance as a gateable, gain-controlled building block for instrumentation and audio applications, such as linearization of transducer outputs, standardization of widely changing signals for data processing, multiplexing, instrumentation amplifiers operating from the nano-power range to high-current and high-speed comparators.

### Features

- Low initial input-offset voltage: 500 $\mu$ V max. (CA3280A)
- Low Offset-voltage change versus  $I_{ABC}$ : <500  $\mu$ V typ. for all types
- Low offset-voltage drift: 5  $\mu$ V/C max. (CA3280A)
- Excellent matching of the two amplifiers for all characteristics
- Internal current-driven linearizing diodes reduce the external input current to an offset component

## CA3060 Variable Op-Amp Array CA3078 Micropower Op-Amp

CA3078: Low standby power assures operation from a single 1.5-V battery; a programmable input terminal permits tailoring the frequency response and slew rate without sacrificing power.

CA3060: 3 variable op-amps on one chip plus a zener bias regulator to control bias current or as a voltage-reference source for external circuits.

Sym- bol	Characteristics Limits, $T_A = 25^\circ\text{C}$						Units
	CA3060 (ea. amplifier)			CA3078A			
	$I_{ABC} = 100 \mu\text{A}$			$R_{SET} = 5.1 \text{ M}\Omega; I_Q: 20 \mu\text{A}$			
	Min.	Typ.	Max.	Min.	Typ.	Max.	
$V_{IO}$	—	1	5	—	1.3	3.5	mV
$I_{IO}$	—	250	1000	—	0.5	2.5	$\mu\text{A}$
$I_{IB}$	—	2500	5000	—	7	12	nA
$I_{OM}$	150	240	—	—	$\pm 12$	—	mA
+	12	13.6	—	5.1	5.3	—	V
$V_{OM}$	—	—	—	—	—	—	
-	12	14.7	—	5.1	5.3	—	V
$V_{ABC}$	—	0.66	—	—	—	—	
$g_{21}$	30	102	—	—	—	—	mmho
CMRR	70	90	—	80	115	—	dB
$V_{ICR}$	+12 to -12	+13 to -14	—	-5 to 5	-5.5 to +5.8	—	V
SR	—	8	—	—	—	—	V/ $\mu$ s

CA3078  $A_{OL}$ : 88 db min., 92 typ. dB

- Differential amplifier emitters brought out for use in emitter-coupled dual-differential amplifier applications
- Low noise: 8 nV/ $\sqrt{\text{Hz}}$  @ 1 kHz typ.
- Low noise distortion: 0.4% THD typ.
- Two modes of gain control

### Recommended Applications

- Voltage-controlled amplifiers
- Voltage-controlled filters
- Voltage-controlled oscillators
- Multipliers
- Demodulators Sample and hold
- Instrumentation amplifiers
- Function generators
- Triangle wave-to-sine wave converters
- Comparators
- Audio preamplifiers

Sym- bol ( $T_A = 25^\circ\text{C}$ )	Characteristics and Limits						Units	
	CA3280			CA3280A				
	Min.	Typ.	Max.	Min.	Typ.	Max.		
$R_i$	0.5	—	—	0.5	—	—	$\text{M}\Omega$	
$I_B$	—	3	8	—	3	8	$\mu\text{A}$	
$I_{IO}$	—	0.3	0.7	—	0.3	0.7	$\mu\text{A}$	
$V_{IO}$	—	0.7	3	—	0.25	0.5	mV	
SR	—	125	—	—	125	—	V/ $\mu$ s	
9m	—	16	22	—	16	22	mmho	
+	$V_{ICR}$	-13	—	13	-13	—	13	V
-								
+	$V_{OM}$	12	13.9	—	12.5	13.9	—	V
-		-12	-14.5	—	-13.5	-14.5	—	
$V_{ABC}$	—	1.2	—	—	1.2	—	V	

### Features

#### CA3078

- Low standby power: as low as 700 nW
- Wide supply-voltage range:  $\pm 0.75$  to  $\pm 15$  V
- High peak output current: 6.5 mA min.
- Adjustable quiescent current
- Output short-circuit protection

#### CA3060

- Low power consumption: as low as 100 $\mu$ W/amplifier
- Independent biasing for each amplifier
- Programmable range of input characteristics
- Low input bias and input offset current
- No effect on device under output short-circuit conditions
- High forward transconductance

### Recommended Applications

#### CA3078

- Portable electronics
- Medical electronics
- Instrumentation
- Telemetry

#### CA3060

- Active filters
- Gytrators
- Multiplexers
- Strobing and gating functions
- Sample-and-hold functions
- Mixers
- Comparators
- Modulators
- Multipliers

# LIC Operational Amplifiers-OTAs

**CA3094 Programmable Power Switch/Amplifier**  
**CA3080 Programmable Op-Amp**

- Input control permits user to vary voltage, power bandwidth, slew rate, input and output currents
- Devices can be programmed and/or signal-modulated for optimum gain, speed, bandwidth, power
- Outputs can "sink" or "source" currents
- Linear gain is adjustable over 3 decades of amplifier bias current  $I_{ABC}$
- Standby power levels can be adjusted from below  $10\mu W$  up to 30 mW

## Features

- CA3094**
- Programmable: strobing, gating, squelching, agc capabilities
  - Capable of delivering 3 W average or 10 W peak to external loads
  - High current-handling capability: 100 mA avg., 300 mA peak
  - Sensitivity controlled by varying bias current
  - Power output of 0.6 W in high-power, single-ended class A amplifier (1.6 W dissipation)
  - THD: 1.4% typ. (at 0.6 W class A)
  - Designed for single or dual power supplies
- CA3080**
- Differential input, single-ended class A output
  - $I_{ABC}$  used for either gating or linear gain control
  - Slew rate (unity gain, compensated): 50 V/ $\mu s$
  - Flexible supply voltage range:  $\pm 2$  V to  $\pm 15$  V
  - Fully adjustable gain: 0 to  $g_m R_L$  limit
  - Tight  $g_m$  spread: 2:1 for CA3080, 1.6:1 for CA3094

## Recommended Applications

- CA3094**
- Fire detection and intrusion alarm systems
  - Portable GFI detectors
  - Function generators
  - Temperature-measuring equipment
  - Driver for complementary power transistors for audio use
  - Battery-operated flashers
  - Power supplies
- CA3080**
- Communications equipment
  - Function generators (1,000,000-to-1 sweep capability)
  - Peak detector instruments
  - Battery-operated paging systems
  - Audio-frequency synthesizers

Sym- bol	Characteristics Limits, $T_A = 25^\circ C$						Units
	CA3094			CA3080			
	$I_{ABC} = 100 \mu A$			$I_{ABC} = 500 \mu A$			
	Min.	Typ.	Max.	Min.	Typ.	Max.	
$V_{IO}$	—	0.4	5	—	0.4	5	mV
$I_{IO}$	—	0.02	0.2	—	0.12	0.6	$\mu A$
$I_{IB}$	—	200	500	—	2	5	$\mu A$
$I_{OM}$	—	—	—	350	500	650	$\mu A$
+ $V_{OM}$	14.95	14.99	—	12	13.5	—	V
	—	14.96	—	12	14.4	—	
$V_{ABC}$	—	0.68	—	—	0.71	—	V
$g_m$	1.65	2.2	2.75	6.7	9.6	13	mmho
CMRR	70	110	—	80	110	—	dB
$V_{ICR}$	+12†	+13.8†	—	+12 to	+13.6	—	V
	-14‡	-4.5‡	—	-12	-14.6	—	
SR	500 (max. value) 0.7 (unity gain)			—	—	75	V/ $\mu s$

†  $V^+ = 15V$  ‡  $V^- = 15V$

# LIC Voltage Comparators

Dual and Single Types

Sym- bol	Characteristics Limits: $T_A = 25^\circ C$					Units
	Dual			Single		
	BiMOS CA3290,A,B			CA311		
	V+	Typ.	Max.	Max.		
$V_{IO}$	+5	7.5	20	—	—	mV
	$\pm 15$	7.5	20	10	—	
$I_{IB}$	+5	3.5 pA	50 pA	—	—	nA
	$\pm 15$	12	50 pA	250	—	
$I_{IO}$	+5	2 pA	30 pA	—	—	nA
	$\pm 15$	7 pA	30 pA	50	—	
$V_{ICR}$	+5†	Typ. Value: $V^+ - 3.5$ to $V^- - 1.5$		—	—	V
	$\pm 15$ ‡	Typ. Value: $V^+ - 3.4$ to $V^- - 1.6$		—	—	
$I_{SINK}$	+5	30	—	—	—	mA
$V_{sat}$	+5	0.12	0.4	—	—	V
$T_R$	$\pm 15$	R	1.2	—	—	$\mu s$
		F	200	—	200*	ns
CMRR	$\pm 15$	44	562	—	—	$\mu V/V$
	+5	100	562	—	—	

† Min. value:  $V^+ - 3.5$  to V; ‡ Min. value:  $V^+ - 3.8$  to V  
R = Rising Edge F = Falling Edge \*Typical  
 $T_R$  = Response Time at  $R_L = 5.1 k\Omega$

## Features

- CA3290, CA3290A, CA3290B**
- State-of-the-art design, first in the industry
  - MOS/FET input/bipolar output BiMOS types
  - Wide  $V_{ICR}$ , can be swung 1.5 V below the negative supply rail, permits single-supply operation
  - Ultra-low input current (3.5 pA typ.), eliminates need for high-impedance buffer circuits
  - Wide differential input voltage range (to  $\pm 36$  V), allows input circuits to operate at high  $V^+$
  - No phase reversal of output signal when input transients go below ground
  - Very low supply current drain (0.8 mA at +5 V)
  - Low output saturation voltage (120 mV at 4 mA)
  - 14-pin package to reduce stray capacitances between input and output circuits

## CA311

- Directly interchangeable with National LM311 series and other 311 industry types

## Recommended Applications

- High-source-impedance voltage comparators
- Long-time-delay circuits
- A/D converters
- Window detector circuits
- Zero-crossing detectors
- Positive and negative peak detectors

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# LIC Voltage Comparators

## Quad and Programmable Types Programmable Types with Memory

Symbol	Characteristics Limits: $T_A = 25^\circ\text{C}$						Units
	CA3098			CA3099			
	Min.	Typ.	Max.	Min.	Typ.	Max.	
$V_{IO(LR)}$	-15	-3	6	-8	-3	2	mV
$V_{IO(HR)}$	-10	$\pm 10$	10	-5	$\pm 1$	5	mV
$V_{IO(HR-LR)}$	-	3	20	-	3	10	mV
Temp. Coeff.	-	6.7	-	-	6.7	20	$\mu\text{V}/^\circ\text{C}$
$V_{CE(\text{sat})}$	-	0.72	1.2	-	0.72	1.2	V
$I_{T\text{ ON}}$	500	710	800	600	710	800	$\mu\text{A}$
OFF	400	560	750	420	560	750	$\mu\text{A}$
$I_{IB}$	-	42	100	-	33	200	nA
$I_{OL(\text{OFF})}$	-	-	10	-	-	10	$\mu\text{A}$
$I_O$	100	-	-	-	-	-	mA
$t_d$	-	600	-	-	600	-	ns
$t_i$	-	50	-	-	50	-	ns
$t_r$	-	500	-	-	500	-	ns
$t_s$	-	4.5	-	-	4.5	-	$\mu\text{s}$
$V_{REF}$	-	-	-	5.7	6	6.3	V
$V_{REG}$	-	-	-	6	7.2	8	V

### Quad Types

Sym- bol	Max. Characteristics Limits: $T_A=25^\circ\text{C}$ , $V^+=+5\text{V}$				Units
	CA139	CA139,A	CA239,A	CA339,A	
$V_{IO}$	5	2	-	-	mV
$I_{IO}$	25	25	50	50	nA
$I_{IB}$	100	100	250	250	nA
$V_{ICR}$	$V^+-1.5$	$V^+-1.5$	-	-	V
$I^*$	2	2	-	-	mA
$I_{OL}$	0.1	0.1	-	-	nA
$I_{OM}$	16*	16*	-	-	mA

## Data Conversion Circuits

### Display Drivers

- CA3207 BiMOS Sequencer Driver
- CA3208 BiMOS Segment Latch-Driver

The RCA CA3207 and CA3208 sequence-driver and segment latch-driver, respectively, are used in combination to drive vacuum fluorescent display devices of up to 14 characters of display. The CA3207 selects the digit or character to be displayed in sequence and the CA3208 turns on the required number of segments of the character selected.

### Features

- Serial input, parallel output
- Total of 14 outputs
- CMOS and T<sup>2</sup>L compatible inputs
- Low-power CMOS Logic-Bipolar high-voltage output BiMOS process
- Use with vacuum fluorescent display
- Will operate in an output voltage range of 35 V to 55 V

#### Sequencer Driver (CA3207E)

- Sequentially turns on 1 of 14 characters (or 2 of 28 when used with 2 CA3208E's)
- Signal dimming through Gates 1 or 2

### Features

- Designed to control high-operating-current loads such as thyristors, lamps, relays, LED's
- Can be operated with single power supply (16 V max.) or dual power supply ( $\pm 8\text{V}$ )
- Directly control currents up to 150 mA
- Operate with microwatt standby power when current to be controlled is less than 30 mA
- Programmable operating current
- High sensor range: 100 ohms to 100 megohms
- Built-in hysteresis: 20 mV max. for CA3098, 10 mV max. for CA3099
- Programmable hysteresis: 20 mV to  $V^+$  for CA3098, 10 mV to  $V^+$  for CA3099
- Temperature-compensated reference voltage (CA3099)

### Recommended Applications

- Signal reconditioning
- On-off motor switching
- Schmitt triggers, level detectors
- Time delays
- Overvoltage, overcurrent, overtemperature protection
- Battery-operated equipment

### Features

- Direct replacements for industry types 139, 139A, 239, 239A, 339, 339A
- Low supply-current drain, suitable for battery operation
- Common-mode input-voltage range to ground

### Recommended Applications

- Square-wave generators
- Pulse generators
- Time-delay generators
- High-voltage digital logic gates
- Multivibrators
- A/D converters
- MOS clock timers

### Latch Driver (CA3208E)

- Drives any combination of 14 outputs selected by DATA input
- Two or more devices may be interconnected by means of the CE and  $\bar{C}E$  inputs to drive more than 14 characters

### CA3168 2-Digit BCD-to-7-Segment Decoder/Driver

The RCA CA3168 is a monolithic integrated circuit intended for 2-digit display such as "numbers" for TV and "CB" channel selection, and other 0-99 numerical or counting for consumer or industrial indicator applications.

### Features

- Separate BCD inputs and segment outputs for each digit
- Input loading less than  $15\mu\text{A}$
- I<sup>2</sup>L logic with buffered inputs and outputs
- Internal input overrange protection circuit
- 5-V supply operation
- Internal biasing circuits
- Output drive capability of 25 mA per segment
- Open collector outputs drive indicators directly

# Data Conversion Circuits

## A/D Converters

- CA3300 CMOS Video Speed 6-Bit Flash A/D Converter**
- CA3308 CMOS Video Speed 8-Bit Flash A/D Converter**

The RCA CA3300 and CA3308 are CMOS parallel (FLASH) A/D converters designed for applications demanding both low power consumption and high-speed digitization. They operate over a wide full-scale input-voltage range with maximum power consumption as low as 50 to 200 mW depending upon the clock frequency selected.

### Features

- CMOS low power with speed
- Parallel conversion technique
- 15 MHz sampling rate (66-ns conversion time)
- $\pm 1/2$  LSB accuracy
- 2 units in parallel allow 30-MHz sampling rate

### Recommended Applications

- The devices are especially suited for high-speed conversion applications where low power is also important
- High-speed A/D conversion
- Ultrasound signature analysis
- Transient signal analysis
- High-energy physics research
- High-speed oscilloscope storage/display
- General-purpose hybrid A/Ds
- Optical character recognition
- Radar pulse analysis
- Motion signature analysis

### Features CA3300

- 6-bit latched 3-state output with overflow bit
- Single supply voltage (3 to 10V)
- 2 units in series allow 7-bit output
- Internal VREF with ext VREF option

### Applications

- TV video digitizing (industrial/security)

- CA3161 BCD-to-Seven-Segment Decoder/Driver**
- CA3162 Converter for 3-Digit Digital Readout System**

The CA3162E analog-to-digital converter is specially designed for use in low-cost digital readout systems. This circuit and a companion type, the CA3161E BCD-to-7-segment decoder/driver, can be used to implement a 2-chip, 3-digit readout system featuring simplicity of operation and a minimum of external parts. The CA3162E provides highly accurate conversions, exceptional temperature stability, and other features listed below.

### Features CA3162E

- Dual-slope A/D conversion for excellent accuracy and noise rejection
- On-chip clock and ultra-stable bandgap reference; no additional IC's required
- Reads 99 mV below ground with single +5 V supply
- Choice of low-speed (4-Hz) or high-speed (96-Hz) conversion rate
- Overrange indication
- Hold function inhibits conversion but maintains display

### CA3161E

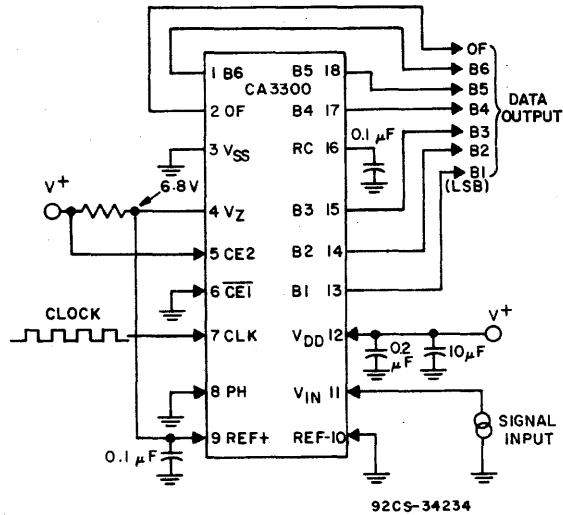
- TTL-compatible input logic levels
- 15-mA typ. constant-current output drive, eliminates need for current-limiting resistors
- Pin-compatible with industry standard decoders

### Features CA3308

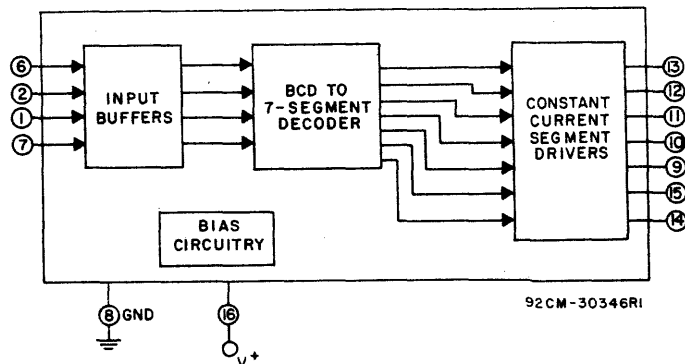
- 8-bit latched 3-state output with overflow bit
- Single supply voltage (4 to 8 V)
- 2 units in series allow 9-bit output

### Applications

- TV video digitizing (industrial/security/broadcast)
- $\mu$ P data acquisition systems



Typical CA3300 6-bit configuration 15-MHz sampling rate.



CA3161 Block Diagram

### Recommended Applications

- Medical equipment
- Welding controls
- Electronic scales
- Electronic games
- Temperature measuring instruments
- Power-supply meters
- Analog panel meters
- Auto Accessories
- Industrial Controls

## LIC Amplifier, Transistor/Diode, and Diode Arrays

### Features

- CA3093 has high  $I_C$  (100 mA max. each transistor) and temperature-compensation capability
- CA3097 has PUT with extremely long time constants, SCR with 150-mA forward current, complete isolation between elements
- CA3019 has excellent diode match; low leakage currents, low pedestal voltage when gating
- CA3039 has ultra-low capacitance diodes: 1 ns typ. reverse recovery time,  $C_D=0.65$  pF typ.,  $V_F$  matched within 5 mV
- CA3141 has  $V_{BR(IR)}$  of 30 V min., low  $I_R$  of 100 nA max., low  $C_D$  of 0.3 pF typ.,  $V_F$  of each diode pair matched to within 0.55 mV at 1-mA  $I_F$
- CA3048, CA3052 include internal dc feedback and bias to provide temperature-stabilized operation

### Recommended Applications

- Lamp and relay drivers, differential amplifiers, temperature-compensated amplifiers
- Timers, light dimmers/motor controls, comparators, Schmitt triggers, constant-current sources
- Analog switches, balanced modulators, mixers, diode gates for chopper modulators
- Balanced modulators or demodulators, ring modulators, high-speed diode gates
- Current-ratio detectors, high-voltage diode gates, analog switches
- Low-level preamplifiers, mixers, tone generators, linear signal mixers

Amplifier Types				Transistor/Diode Types		Diode types		
Dual independent Differential	4 AC Amplifiers	3 Variable Operation Amplifiers	Quad Operational Amplifier	3 n-p-n Transistors 2 Zener Diodes 1 Diode	1 n-p-n, 1 p-n-p/n-p-n pair, 1 Zener Diode, 1 PUT 1 SCR	1 Quad Plus 2	6 Ultra-Fast Low Capacitance	10 High-Voltage (5 Pairs)
CA3026 CA3049# CA3102#	CA3048 CA3052	CA3060 CA3060A	CA124 CA224 CA324	CA3093	CA3097	CA3019	CA3039	CA3141
CA3054	2 Darlington Differential Amplifiers with Diode Bias String CA3050, CA3051							
#High Frequency								

## LIC Transistor Arrays

### Features

- CA3081, CA3082 directly drive 7-segment displays
- CA3127, CA3102, CA3049 are useful in low-power applications up to 500 MHz; low 1/f noise;  $f_T$  greater than 1 GHz
- CA3227, CA3246 consist of five general-purpose silicon n-p-n transistors on a common monolithic substrate,  $f_T$  greater than 3GHz
- CA3600 (PMOS and NMOS devices) is specified and tested for linear circuit applications

Each transistor can conduct currents up to 10 mA.  
High input resistance (100 GΩ typ.).

### Recommended Applications

- Drivers for incandescent display devices, relay controls, thyristor firing
- VHF amplifiers, mixers, rf mixers/oscillators, doubly balanced modulators/demodulators
- Core-memory drivers, high-speed switching, high-current LED drivers, lamp drivers
- For complementary-symmetry circuits at supply voltages of 3 to 15 V and frequencies up to 5 MHz (untuned)
- High-input-impedance, general-purpose amplifiers, preamplifiers, differential amplifiers
- IF converter, IF amplifier, sense amplifiers, synthesizers, synchronous detectors, cascade amplifiers

High-Current		Bipolar Types (▲ High-Beta Types)						CMOS Type	
		High-Voltage			General-Purpose				
7 n-p-n Common Collector	CA3082	5 Independent-3 n-p-n 2 p-n-p	5 Independent n-p-n	Darlington-Connected Pair Plus 2 Independent n-p-n	Differentially Connected Pair Plus 3 Independent n-p-n	Darlington-Connected Pair Plus 2 Independent n-p-n	Differentially Connected Pair Plus 3 Independent n-p-n	5 Independent n-p-n	3 n-Channel and 3 p-Channel
7 n-p-n Common Emitter	CA3081								
5 Independent n-p-n	CA3083 CA3183 CA3183A	CA3096 CA3096A CA3096C	CA3183 CA3183A	CA3118 CA3118A	CA3146 CA3146A	CA3018 CA3018A	CA3045 CA3046 CA3086	CA3227 CA3246	CA3600
4 Isolated n-p-n	CA3138▲ CA3138A▲ CA1724 CA1725	High-Frequency CA3127			1 Darlington Pair, 1 Current Mirror Pair, 2 p-n-p, 1 Shared Diode CA3084				

## LIC Power Control Circuits

### Automotive Ignition Switch

- **CA3165 Electronic Switching Circuit**

The RCA CA3165 is a single-chip electronic switching circuit intended primarily for ignition applications. It includes an oscillator that is amplitude-modulated by the rotor teeth of a distributor, a detector that develops the positive-going modulation envelope, a Schmitt trigger that eliminates switching uncertainties.

#### Features

- Switching initiated by damping of internal oscillator
- Proximity sensing of rotational motion
- Repeatable timing of switching states
- Five outputs — two complementary pairs and one non-inverting output (CA3165E1)
- Two outputs — one complementary pair (CA3165E)

### Solenoid & Motor Driver

- **CA3169 Solenoid and Motor Driver**

The RCA CA3169 is a monolithic integrated circuit capable of driving lamps and other devices that can be changed between two states (on or off). Transistors, SCRs, and triacs are some of the solid-state devices that can be controlled by the CA3169. The device also can control relays, solenoids (latching or non-latching), motors (DC-forward and reverse) and DC stepping motors.

#### Features

- Chip encapsulated in a 5-lead plastic TO-220-style package (VERSA-VI)
- Output short-circuit protection
- Thermal overload protection
- Solenoid inductive "kick" protection with internal-clamp diodes
- Output sink and source capacity of 600-mA minimum overtemperature
- Horizontal and vertical mounting packages available.
- Separate sink circuit and source circuit, each individually controlled
- Inputs can be driven by TTL logic levels and CMOS logic levels
- Low  $V_{CE(sat)}$

#### Applications

- Latching solenoid driver (single and multiple)
- Non-latching solenoid driver
- Relay driver
- Lamp controller
- Lamp driver
- Motor controller (forward and reverse)
- Stepper motor controller
- On-off logic controllers (TTL logic)
- Intermediate power driver
- Triac, SCR, and transistor drivers

- **CA3219 Quad-Power NAND Driver**

The RCA CA3219 quad power NAND driver contains four NAND gate switches for interfacing low-level logic to inductive and resistive loads such as: relays, solenoids, AC and DC motors, heaters, incandescent displays, and vacuum fluorescent displays.

#### Features

- Driven outputs capable of switching 600 mA load currents without spurious changes in output state
- Inputs compatible with TTL or 5-volt CMOS logic
- Suitable for resistive or inductive loads

### Universal Controller

- **CA3228 Speed-Control System**

The RCA CA3228 is a monolithic  $\text{I}^2\text{L}$  integrated circuit designed as an automotive speed-control system. The system monitors vehicle speed and compares it to a set reference speed. Any deviation in vehicle speed causes a servo mechanism to open or close the engine throttle as required to eliminate the speed error. The reference speed is set by the driver to hold the existing speed and stored in a 9-bit counter.

#### Features

- Low power dissipation
- $\text{I}^2\text{L}$  control logic
- Power-ON reset
- On-chip oscillator for system time reference
- Single line command
- Amplitude encoded control signals
- Transient compensated input commands
- Controlled acceleration mode
- Internal redundant brake and low speed disable
- Braking disable

### Voltage Regulators

- **CA723, CA723C**

- Up to 150 mA output current
- Regulation in excess of 10 A with suitable pass transistors
- Input and output short-circuit protection
- Load and line regulation of 0.03%
- Adjustable output voltage of 2 V to 37 V
- Direct replacements for 723, 723C industry types

- **CA3085, CA3085A, CA3085B**

- Up to 100 mA output current
- Regulation up to 12 mA without pass transistors, in excess of 100 mA with suitable pass transistors
- Load and line regulation of 0.025%
- Low noise • Frequency compensation
- Input and output short-circuit protection
- Temperature-compensated reference voltage
- Pin-compatible with LM100 series

#### Recommended Applications

- Series and shunt voltage regulators
- Current regulators • Floating regulators
- Temperature controllers
- High-current voltage regulators
- Dual-tracking regulation
- Combination positive and negative voltage regulators

- **CA1524, CA2524, CA3524**

- Complete PWM power control Circuitry
- Separate outputs for Single-ended or push-pull operation
- Line and load regulation of 0.2% typ.
- Internal reference supply with 1% max. oscillator and reference voltage variation over full temperature range
- Standby current of less than 10 mA
- Frequency of operation beyond 100kHz
- Variable output dead time of 0.5 to 5  $\mu\text{s}$
- Low  $V_{CE(sat)}$  over the temperature range

#### Recommended Applications

- Positive and negative regulated supplies
- Dual-output regulators
- Flyback converters
- DC-DC transformer-coupled regulating converters
- Single-ended DC-DC converters
- Variable power supplies

## LIC Power Control Circuits

### Zero-Voltage Switches

#### Features

CA3058, CA3059, CA3079

- Operation at 24 V, 120 V, 208/230 V, 227 V at 50, 60, or 400 Hz
- Differential on-off sensing amplifier
- Zero-crossing detector • Triac gating circuit
- Low balanced input current: 1  $\mu$ A max.
- Built-in protection circuit for open or shorted sensor
- Differential input
- Sensor range of 2 to 100 kilohms

### Recommended Applications

CA3058, CA3059, CA3079

- Relay control • Valve control • Lamp control
- Synchronous switching of flashing lights
- On-off motor switching
- Differential comparators with self-contained power supply for industrial applications
- Photosensitive controls
- Power one-shot controls

## LIC Differential Amplifiers

#### Features

- Balanced differential-amplifier configuration with controlled constant-current source
- RF, if, and video frequency capability
- Balanced agc capability
- Operation from dc to 120 MHz
- CA3028B is controlled for input offset voltage, current, and input bias current, and is intended for "balance" requirements
- CA3005 and CA3006 are identical except for input offset voltage

#### Applications

- RF amplifiers • Oscillators
- Mixers • Limiters
- IF amplifiers • Product detectors

- Frequency generators • Pulse amplifiers
- Preamplifiers for low-level transducers
- DC, audio, and sense amplifiers

T <sub>A</sub> = 25°C; V* = 6V Unless Otherwise Specified							
Type	V <sub>io</sub> Max. mV	I <sub>io</sub> Max. $\mu$ A	I <sub>ib</sub> Max. $\mu$ A	P <sub>D</sub> Typ. mW	G <sub>p</sub> # Typ. dB	NF# Typ. dB	CMR† Typ. dB
CA3005	5	1.4*	40	26	20	7.8	101
CA3006	1	1.4*	40	26	20	7.8	101
CA3028A	—	—	70	36	20†	7.2†	—
CA3028B	5	5	40	36	20†	7.2†	110
CA3053	—	—	—	50†	39*	—	—
*Typical #f 100 MHz †f 1.75 MHz †V* 9 V							
*V* 9 V, f 10.7 MHz							

## LIC Special Function Circuits

### Prescalers

CA3179E

The RCA-CA3179E is an integrated-circuit prescaler intended for use in communications and instrumentation systems, it performs division by 256 in the uhf mode and division by 64 in the vhf mode.

#### Features

- Broadband operation — DC to 1.25 GHz
- High sensitivity
- Standard T<sup>2</sup>L or ECL power supply
- Dual mode operation — VHF/UHF (+64/+256)
- Complementary ECL outputs
- Independent VHF and UHF input terminals

CA3199E

The CA3199E is a bipolar integrated fixed-ratio (divide-by-four) counter which operates over the VHF/UHF frequency band (DC to 1.3 GHz). It accepts either single or double-ended ac-coupled input signals and provides complementary emitter follower outputs at standard ECL logic levels.

#### Features

- Broadband operation — DC to 1.3 GHz
- High sensitivity
- Standard TTL or ECL power supply of 5V  $\pm$  0.5%
- Complementary ECL outputs

### Applications CA3179E, CA3199E

- Digital frequency synthesizers for: VHF/UHF receivers, Satellite communications, instrumentation
- High-frequency divider for: UHF frequency counters, UHF timers, High-speed computers, Frequency standards, SHF second IF local-oscillator injection, PCM communications, Satellite communications, Radar ranging systems
- High-frequency up-converters

CA3211

The RCA-CA3211E is an integrated-circuit prescaler intended for use in TV frequency synthesis tuning systems over an input frequency range of 90 to 1000 MHz. It performs division by 256 in the UHF and VHF mode.

#### Features

- Divide by 256
- Input frequency to 1 GHz
- Dual input ports electrically selectable
- Input sensitivity <10  $\mu$ W typical (Generator available power into a 50- $\Omega$  load)
- 5-V power supply
- Balanced output ports

# LIC Special Function Circuits

## Four Quadrant Multiplier

CA3091

Four-quadrant multiplier that provides an output voltage that is the product of two input (x and y) voltages

### Features

- Accuracy of  $\pm 4\%$  max.
- Internal bias regulator
- Linearity of 3% max.
- 3-dB bandwidth of 4.4 MHz
- Low-power operation capability:  $\pm 6$  V, 4-mW drain
- Low power-supply sensitivity: 36 mV/V typ.
- Functions as a multiplier, divider, squarer, square-rooter, power-series approximator

### Applications

- Ideal full-wave rectifiers
- Automatic level controllers
- Frequency discriminators
- RMS converters
- Voltage-controlled filters and oscillators

## Single-Chip Detector Alarm Systems

CA3164E

The RCA CA3164 is a monolithic BiMOS integrated circuit designed to meet the stringent system requirements of a battery- or line-operated alarm circuit. When used with an ionization chamber and electromechanical or piezoelectric horn, it provides a one-chip approach to smoke detection.

### Features

- Interfaces directly with high Z sensors — no external buffer FET required
- Low input current: 1 pA max.
- Gate-protected input terminals
- On-chip beep oscillator for low battery indication
- Self-contained low-battery-voltage detection circuit
  - (a) Fixed or adjustable trip point available
  - (b) Dynamic battery test when filter capacitor =  $2\mu\text{F}$
- Chamber trigger voltage independent of battery supply voltage (less than 150 mV over temperature and supply variations)

- Reference source current available =  $5\mu\text{A}$  (typ.)
- Low standby battery current =  $8\mu\text{A}$  (typ.)
- Can be used with photoelectric sensors by using a minimum of external passive components in combination with the RCA-CA3078 micropower op-amp
- Multiple-unit interconnect terminal controls a common annunciator circuit
  - (a) A fault to ground doesn't prevent local operation
  - (b) The low battery alarm signal triggers only the local unit
- LED output indicates status of smoke-detector circuit
- Operates from 11 V (max.) supply (either battery or line)
- Battery reversal protection feature

## Timer

CA555, CA555C Types

The RCA-CA555 and CA555C are highly stable timers for use in precision timing and oscillator applications. As timers, these monolithic integrated circuits are capable of producing accurate time delays for periods ranging from microseconds through hours.

### Features

- Accurate timing from microseconds through hours
- Astable and monostable operation
- Adjustable duty cycle
- Output capable of sourcing or sinking up to 200 mA
- Output capable of driving TTL devices
- Normally ON and OFF outputs
- High-temperature stability —  $0.005\%/^{\circ}\text{C}$
- Directly interchangeable with SE555, NE555, MC1555, and MC1455

### Applications

- Precision timing
- Sequential timing
- Time-delay generation
- Pulse generation
- Pulse-width and position modulation
- Pulse detector

## TV/CATV Circuits

AFT	HORIZONTAL/VERTICAL SYSTEMS	MULTIPLEX DECODERS	SOUND IF	TUNING
CA3064 CA3139	CA920A CA1391 CA1394 CA3154 CA3159 CA3202 CA3210 CA3223***	CA758 CA1310A CA3195	CA1190 CA1191 CA2111A CA2136A CA3011 CA3012 CA3013 CA3014 CA3042 CA3065 CA3134	CA3140 CA3152 CA3163 CA3166 CA3168 CA3199 CA3211
CHROMA SYSTEMS	SYNC/AGC CIRCUITS	PIX IF	VIDEODISC CIRCUITS	
CA1398 CA3070 CA3071 CA3072 CA3121 CA3126 CA3128** CA3137 CA3145 CA3151 CA3158 CA3170 CA3172 CA3194** CA3201 CA3217 CA3221	CA3120 CA3142	CA270 CA1352 CA3068 CA3136 CA3153 CA3191 CA3192 CA7607 CA7611	CA2111A CA3215 CA3216 CD3226A@	
@CMOS types, *BiMOS types, **PAL types, ***625 line				

RCA LINEAR

12-BIT MULTIPLYING DIGITAL-TO-ANALOG CONVERTER

Am6012

Preliminary

DESCRIPTION

The Am6012 12-Bit multiplying Digital-to-Analog converter provides high speed and 0.025% differential nonlinearity over its full commercial temperature range.

The D/A converter uses a 3-bit segment generator for the MSBs in conjunction with a 9-bit R-2R diffused resistor ladder to provide 12-bit resolution without costly trimming processes. This technique guarantees a very uniform step size (up to  $\pm 1/2$  LSB from the ideal), monotonicity to 12 bits and integral nonlinearity to 0.05% at its differential current outputs.

The dual complementary outputs of the Am6012 increase its versatility, and effectively double the peak-to-peak output swing. Digital inputs, in addition, can be configured to accept all popular logic families.

While the device requires a reference input of 1mA for a 4mA full scale current, operation is nearly independent of power supply voltage shifts. The power supply rejection ratio is  $\pm 0.001\%FS/\% \Delta V$ . The devices will work from +5, -12V to  $\pm 18V$  rails, with as low as 230mW power consumption typical.

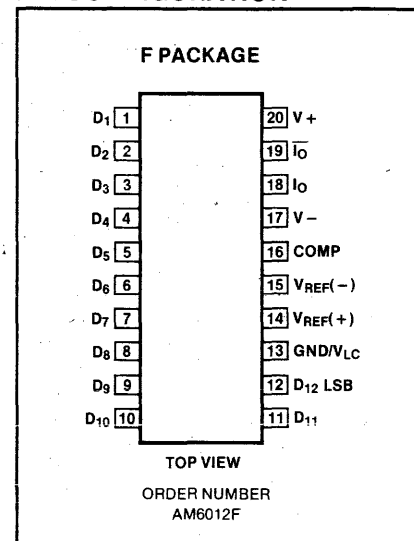
FEATURES

- 12-bit resolution
- Accurate to within  $\pm 0.025\%$
- Monotonic over temperature
- Fast settling time, 250ns typical
- Trimless design for low cost
- Differential current outputs
- High-speed multiplying capability
- Full scale current, 4mA (with 1mA reference)
- High output compliance voltage, -5 to +10V
- Low power consumption, 230mW

APPLICATIONS

- CRT displays, computer graphics
- Robotics, and machine tools
- Automatic test equipment
- Programmable power supplies
- CAD/CAM systems
- Data acquisition and control systems
- Analog-to-Digital converter systems

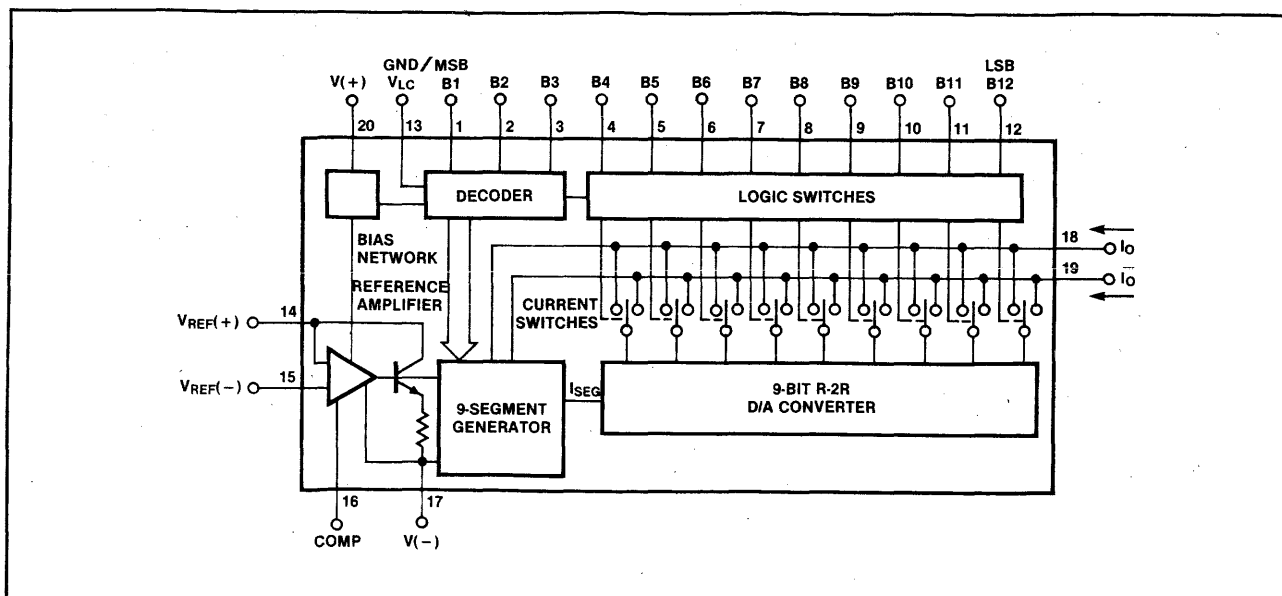
PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS

Operating Temperature Am6012F	0°C to +70°C
Storage Temperature	-65°C to +150°C
Lead Temperature (Soldering, 60 sec)	300°C
Power Supply Voltage	$\pm 18V$
Logic Inputs	-5V to +18V
Voltage Across Current Outputs	-8V to +12V
Reference Inputs $V_{14}, V_{15}$	V- to V+
Reference Input Differential Voltage ( $V_{14}$ to $V_{15}$ )	$\pm 18V$
Reference Input Current ( $I_{14}$ )	1.25mA

BLOCK DIAGRAM



**8-BIT HIGH SPEED MULTIPLYING D/A CONVERTER**

**DAC-08 SERIES**

FORMERLY: NE5007/5008-F,N  
SE5008-F

**DESCRIPTION**

The DAC-08 series of 8-bit monolithic multiplying Digital-to-Analog Converters provide very high speed performance coupled with low cost and outstanding applications flexibility.

Advanced circuit design achieves 85ns settling times with very low glitch and at low power consumption. Monotonic multiplying performance is attained over a wide 20 to 1 reference current range. Matching to within 1 LSB between reference and full scale currents eliminates the need for full scale trimming in most applications. Direct interface to all popular logic families with full noise immunity is provided by the high swing, adjustable threshold logic inputs.

Dual complementary outputs are provided, increasing versatility and enabling differential operation to effectively double the peak-to-peak output swing. True high voltage compliance outputs allow direct output voltage conversion and eliminate output op amps in many applications.

All DAC-08 series models guarantee full 8-bit monotonicity and linearities as tight as 0.1% over the entire operating temperature range are available. Device performance is essentially unchanged over the  $\pm 4.5V$  to  $\pm 18V$  power supply range, with 33mW power consumption attainable at  $\pm 5V$  supplies.

The compact size and low power consumption make the DAC-08 attractive for portable and military/aerospace applications.

**FEATURES**

- Fast settling output current—70ns
- Full scale current prematched to  $\pm 1$  LSB
- Direct interface to TTL, CMOS, ECL, HTL, PMOS
- Relative accuracy to 0.1% maximum over temperature range
- High output compliance -10V to +18V
- True and complemented outputs
- Wide range multiplying capability
- Low FS current drift— $\pm 10$ ppm/ $^{\circ}C$
- Wide power supply range— $\pm 4.5V$  to  $\pm 18V$
- Low power consumption—33mW at  $\pm 5V$

**APPLICATIONS**

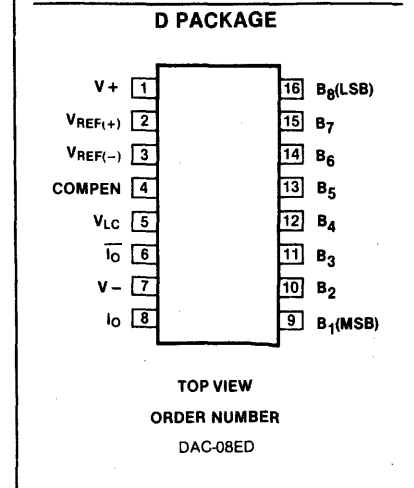
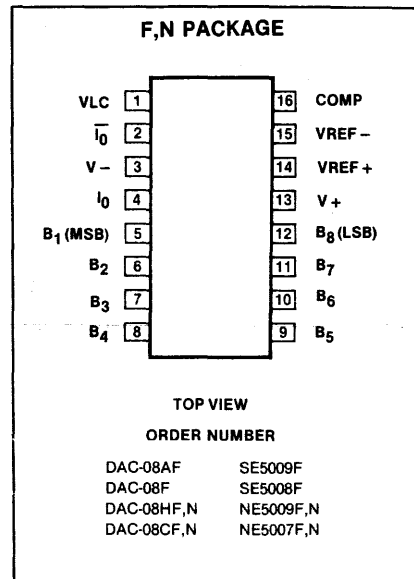
- 8-bit,  $1\mu s$  A-to-D converters
- Servo-motor and pen drivers
- Waveform generators
- Audio encoders and attenuators
- Analog meter drivers
- Programmable power supplies
- CRT display drivers
- High speed modems
- Other applications where low cost, high speed and complete input/output versatility are required
- Programmable gain and attenuation
- Analog-Digital Multiplication
- Stepping motor drive

**ORDERING INFORMATION**

RELATIVE ACCURACY	0 to 70 $^{\circ}C$	-55 to 125 $^{\circ}C$
0.39% FS	DAC-08CN	DAC-08CF
0.19% FS	DAC-08EN	DAC-08EF
	DAC-08ED	DAC-08F
0.1% FS	DAC-08HF	DAC-08AF
	DAC-08HN	

T<sub>A</sub> = 25 $^{\circ}C$  unless otherwise noted

**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
T <sub>A</sub>	Operating temperature range DAC-08, DAC-08A DAC-08C, E, H	$^{\circ}C$
t <sub>stg</sub>	Storage temperature	$^{\circ}C$
P <sub>D</sub>	Power dissipation	mW
	Lead soldering temperature (60sec)	$^{\circ}C$
	V <sub>+</sub> to V <sub>-</sub> supply	V
V <sub>LC</sub>	Logic inputs Logic threshold control Analog current outputs	V <sub>-</sub> to V <sub>-</sub> plus 36V V <sub>-</sub> to V <sub>+</sub> See output current or output voltage performance curve
V <sub>14</sub> , V <sub>15</sub>	Reference inputs	V <sub>-</sub> to V <sub>+</sub>
V <sub>14</sub> to V <sub>15</sub>	Reference input differential voltage	$\pm 18$
I <sub>14</sub>	Reference input current	5.0

LINEAR

Signetics



**8-BIT  $\mu$ P-COMPATIBLE D/A CONVERTER**

**NE/SE5019**

**DESCRIPTION**

The NE5019 is a complete 8-bit digital to analog converter subsystem on one monolithic chip. The data inputs have input latches, controlled by a latch enable pin. The data and latch enable inputs are ultra-low loading for easy interfacing with all logic systems. The latches appear transparent when the  $\overline{LE}$  input is in the low state. When  $\overline{LE}$  goes high, the input data present at the moment of transition is latched and retained until  $\overline{LE}$  again goes low. This feature allows easy compatibility with most micro-processors.

The chip also comprises a stable voltage reference (5V nominal) and a high slew rate buffer amplifier. The voltage reference may be externally trimmed with a potentiometer for easy adjustment of full scale, while maintaining a low temperature co-efficient.

The output of the buffer amplifier may be offset so as to provide bipolar as well as unipolar operation.

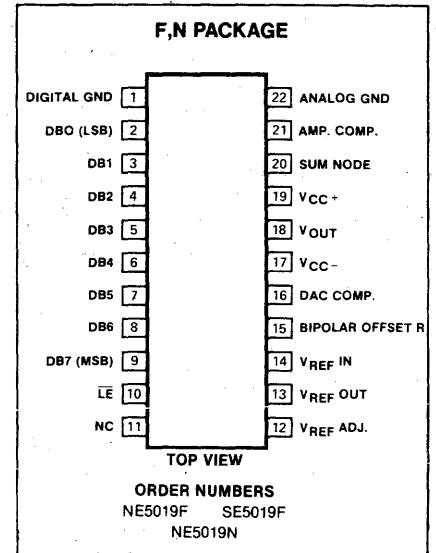
**FEATURES**

- 8-bit resolution
- Input latches
- Low-loading data inputs
- On-chip voltage reference
- Output buffer amplifier
- Accurate to  $\pm 1/4$  LSB (.1%)
- Monotonic to 8 bits
- Amplifier and reference both short-circuit protected
- Compatible with 8085, 6800 and many other  $\mu$ P's

**APPLICATIONS**

- Precision 8-bit D/A converters
- A/D converters
- Programmable power supplies
- Test equipment
- Measuring instruments
- Analog-digital multiplication

**PIN CONFIGURATION**



**BLOCK DIAGRAM**

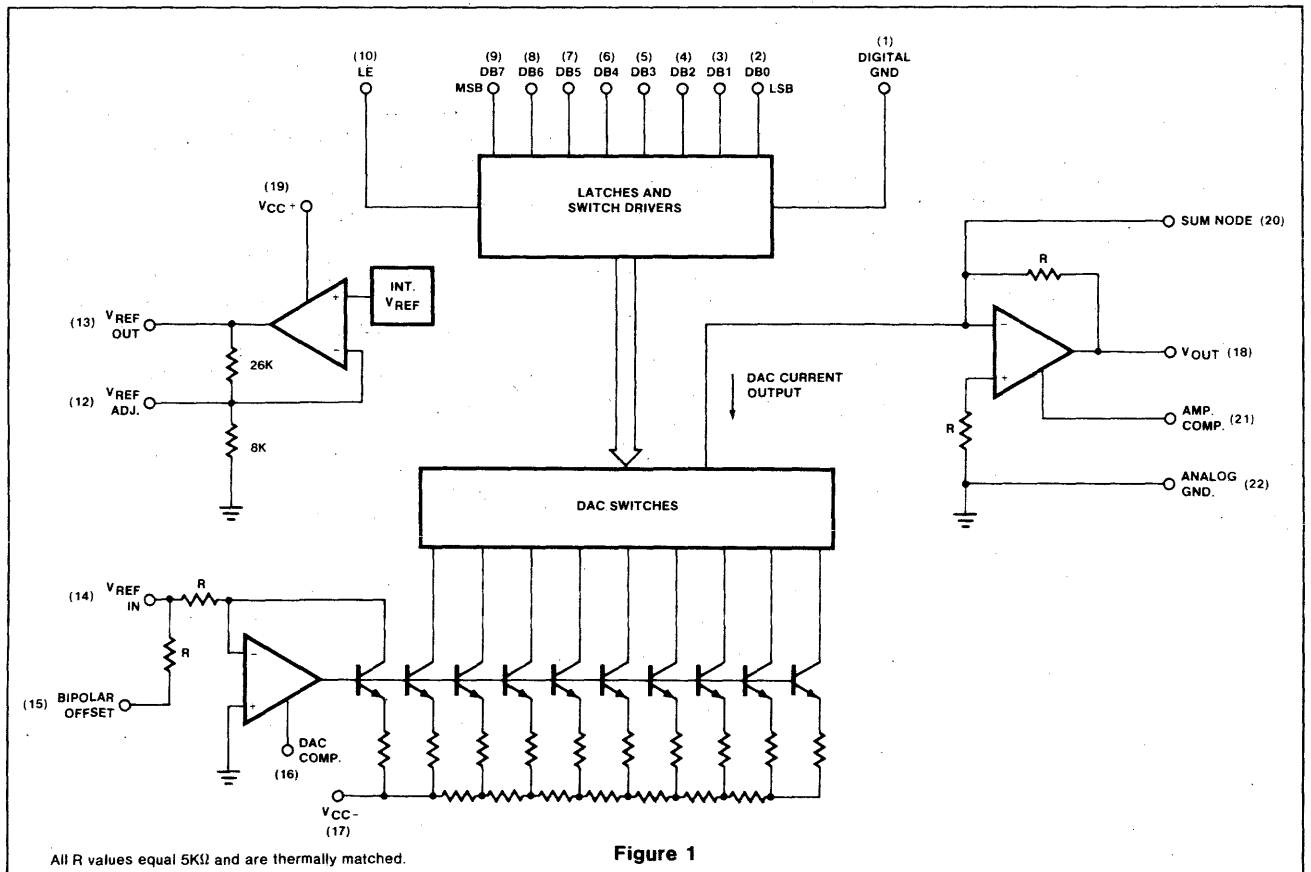


Figure 1

**10-BIT  $\mu$ P-COMPATIBLE D/A CONVERTER**

**NE5020**

**DESCRIPTION**

The NE5020 is a microprocessor-compatible monolithic 10-bit digital to analog converter subsystem. This device offers 10-bit resolution and  $\pm 0.1\%$  accuracy and monotonicity guaranteed over full operating temperature range.

Low loading latches, adjustable logic thresholds and addressing capability allow the NE5020 to directly interface with most microprocessor and logic controlled systems.

The NE5020 contains internal voltage reference, DAC switches and resistor ladder. Also, the input buffer and output summing amplifier are included. In addition, the matched application resistors for scaling either unipolar or bipolar output values are included on a single monolithic chip.

The result is a near minimum component count 10-bit resolution DAC system.

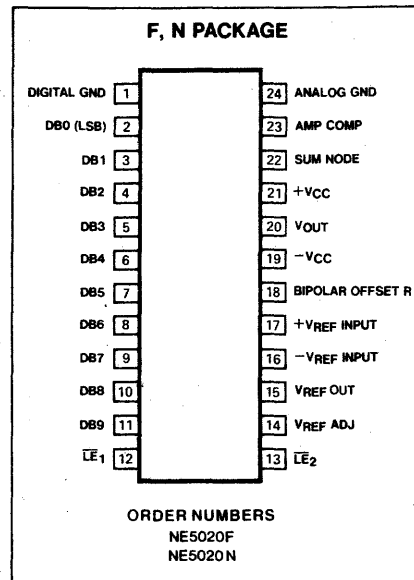
**FEATURES**

- 10-bit resolution
- Guaranteed monotonicity over operating range
- $\pm 0.1\%$  relative accuracy
- Unipolar (0V to +10V) and Bipolar ( $\pm 5V$ ) output range
- Logic bus compatible
- 5 $\mu$ sec settling time

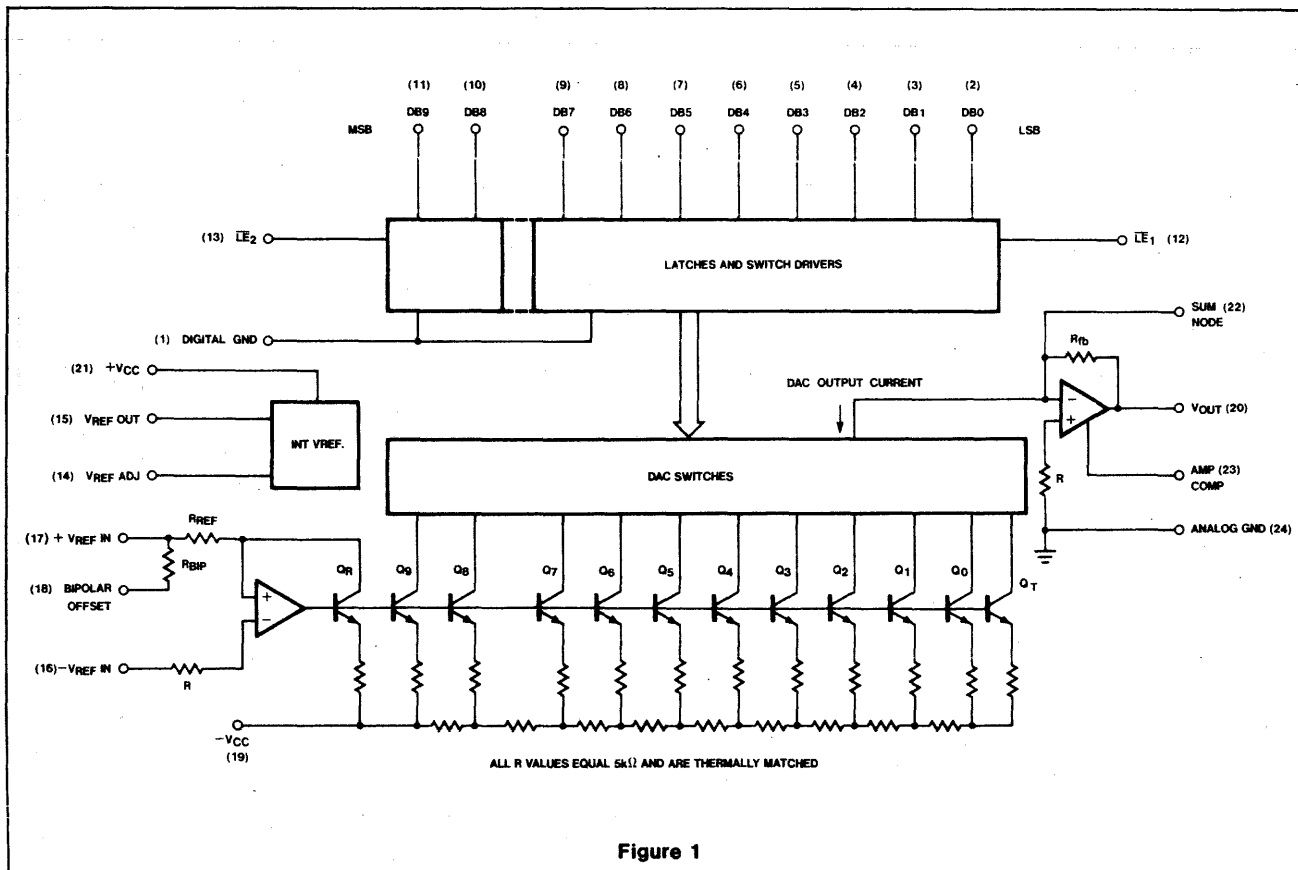
**APPLICATIONS**

- Precision 10-bit D/A converters
- 10-bit Analog to Digital converters
- Programmable power supplies
- Test equipment
- Measurement instruments

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



LINEAR

Signetics

**8-BIT GENERAL PURPOSE A/D CONVERTER**

**NE5034**

**DESCRIPTION**

The NE5034 is a high-speed microprocessor-compatible 8-bit Analog-to-Digital converter. It uses the successive approximation conversion technique, and includes the comparator, reference DAC, SAR, an internal clock and three-state buffers all on the same chip.

The converter can accommodate a wide analog input voltage range, bipolar or unipolar, selectable through external input resistors. An external capacitor controls the internal clock frequency, providing conversion times down to 17µs. Faster conversion times are possible using an external clock.

Microprocessor interfacing requirements are simple, allowing analog-to-digital conversion with a minimum of external components.

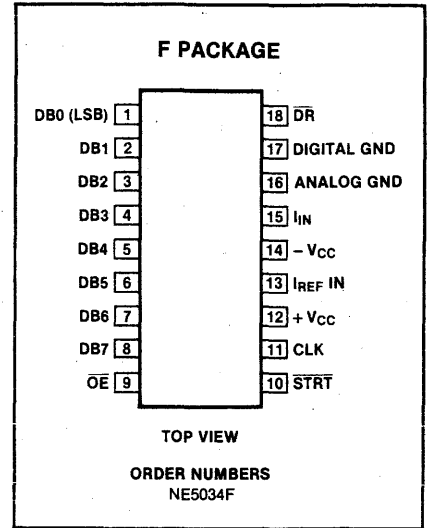
**FEATURES**

- 8-bit resolution and accuracy
- Accepts unipolar or bipolar inputs
- Three-state output buffers for easy microprocessor interface
- Choice of internal or external clocking
- Short conversion time, 17µs typical using internal clock

**APPLICATIONS**

- All microprocessor-based monitoring and control systems requiring analog signal inputs.
- Typical applications include: Automated process control, machine tools, robots, test and measurement instruments, environmental controls
- Other applications include: Ratiometric A/D conversion, very high resolution A/D conversion systems requiring high speed 8-bit building blocks

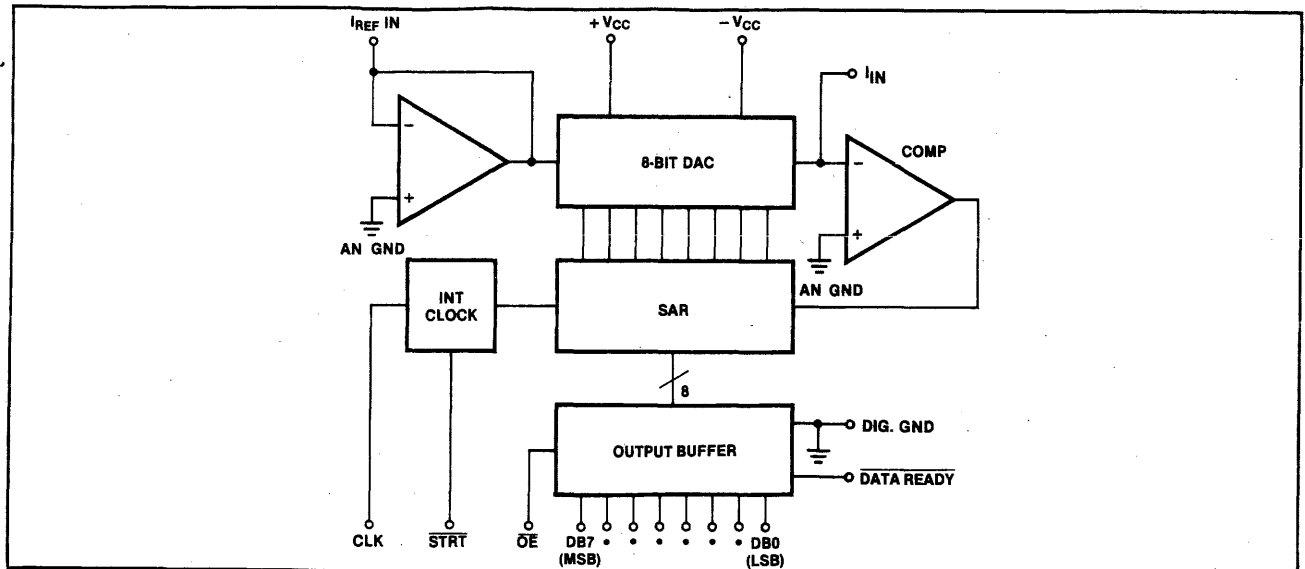
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC+</sub> Positive supply voltage	0 to +6	V
V <sub>CC-</sub> Negative supply voltage	0 to -15	V
I <sub>REF</sub> Reference current	1.5	mA
I <sub>IN</sub> Analog input current	5.0	mA
V <sub>O</sub> Data output voltage	6.0	V
Analog GND to Digital GND	1.0	V
V <sub>L</sub> Logic input voltage	-1 to V <sub>CC+</sub>	V
P <sub>D</sub> Power dissipation		
F package	1000	mW
T <sub>A</sub> Operating temperature range	0 to +70	°C
T <sub>STG</sub> Storage temperature range	-65 to +150	°C
T <sub>SOLD</sub> Lead soldering temperature (10 seconds)	300	°C

**BLOCK DIAGRAM**



**6-BIT A/D CONVERTER (PARALLEL OUTPUTS)**

**NE5037**

**DESCRIPTION**

The NE5037 is a low cost, complete successive approximation analog to digital (A/D) converter, fabricated in Bipolar/1<sup>2</sup>L technology. With an external reference voltage, the NE5037 will accept input voltages between 0V and V<sub>REF</sub>. An external START pulse of at least 300ns in duration will provide the 6-bit result of the conversion in parallel format. Full conversion with no missing codes occurs in 9μs.

**FEATURES**

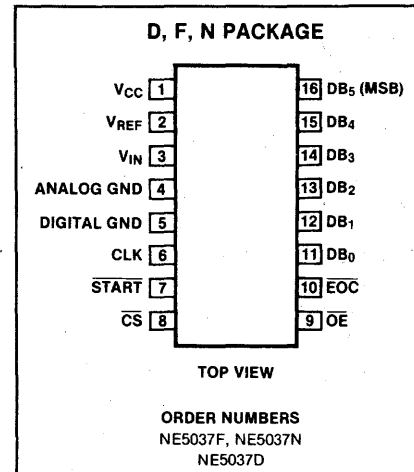
- T<sup>2</sup>L compatible inputs and outputs
- Three state output buffer

- Easy interface to CMOS μProcessors
- Fast conversion—9μs
- Guaranteed no missing codes over full temp range
- Single supply operation, +5V
- Positive true binary outputs
- High impedance analog inputs

**APPLICATIONS**

- Temperature control
- μP-based appliances
- Light level monitors
- Head position sensing
- Electronic toys
- Joystick interface

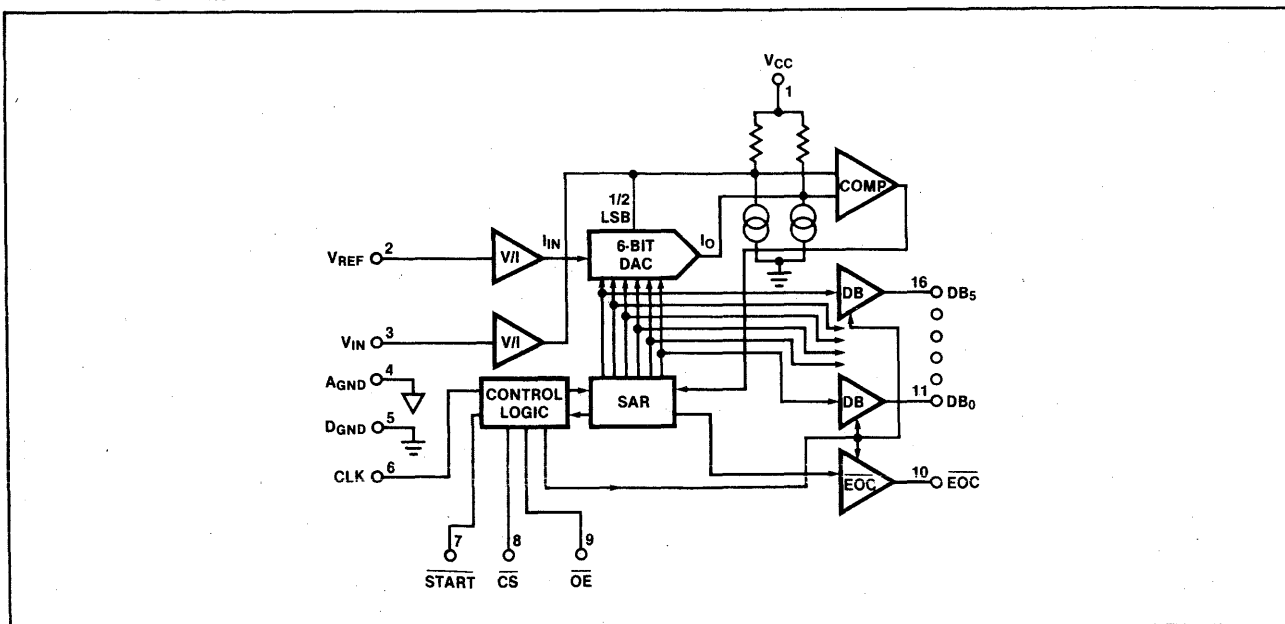
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub>	7	V
V <sub>REF</sub>	7	V
V <sub>IN (Analog)</sub>	7	V
V <sub>IN (Digital)</sub>	7	V
D <sub>OUT</sub>	7	V
	5	mA
EOC	V <sub>CC</sub>	V
ΔGND	±1	V
T <sub>A</sub>	0 to 70	°C
T <sub>STG</sub>	-65 to 150	°C
t <sub>SOLD</sub>	300	°C
P <sub>D</sub>	220	mW
	220	mW

**BLOCK DIAGRAM**



LINEAR

Signetics

**8-BIT  $\mu$ P-COMPATIBLE D/A CONVERTER — CURRENT OUTPUT**

**NE/SE5119**

**DESCRIPTION**

The NE5119 is a high-speed 8-bit digital to analog converter subsystem on one monolithic chip. The data inputs have input latches, controlled by a latch enable pin. The data and latch enable inputs are ultra-low loading for easy interfacing with all logic systems. The latches appear transparent when the LE input is in the low state. When LE goes high, the input data present at the moment of transition is latched and retained until LE again goes low. This feature allows easy compatibility with most micro-processors.

The chip also comprises a stable voltage reference (5V nominal). The voltage reference may be externally trimmed with a potentiometer for easy adjustment of full scale, while maintaining a low temperature co-efficient.

The output has high voltage compliance increasing versatility.

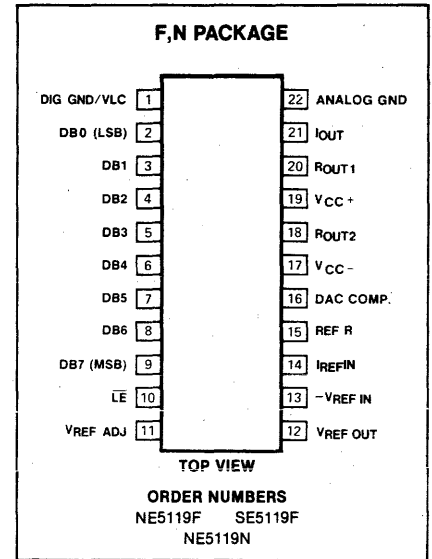
**FEATURES**

- 8-bit resolution
- Input latches
- Low-loading data inputs
- On-chip voltage reference
- Fast settling output current—200ns
- Accurate to  $\pm 1/4$  LSB (.1%)
- Monotonic to 8 bits
- Reference short-circuit protected
- Compatible with 8086, 6800 and many other  $\mu$ P's

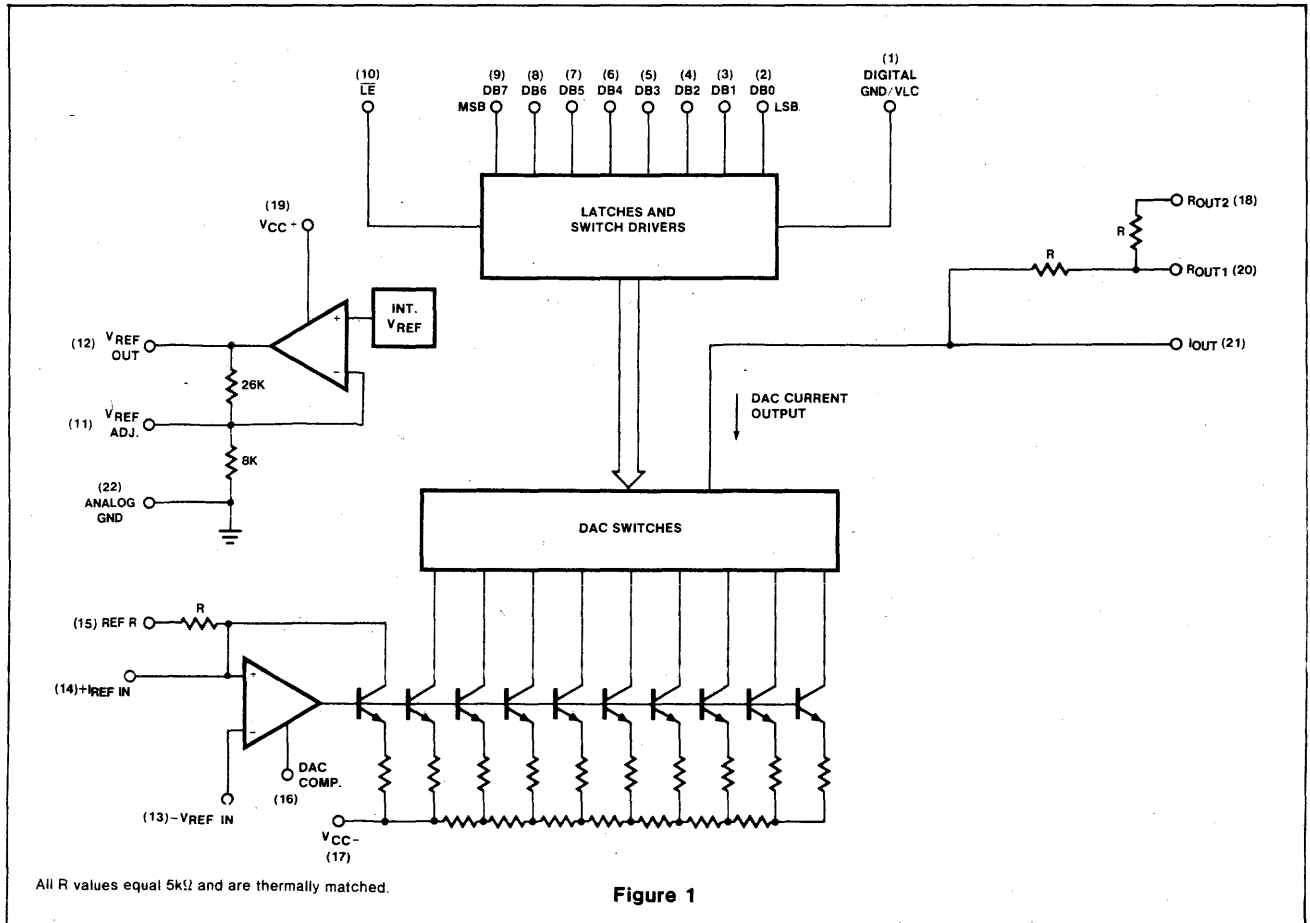
**APPLICATIONS**

- Precision 8-bit D/A converters
- A/D converters
- Programmable power supplies
- Test equipment
- Measuring instruments
- Analog-digital multiplication
- CRT display drivers
- High-speed modems

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**10-BIT HIGH-SPEED MULTIPLYING D/A CONVERTER**

**NE/SE5410**

**DESCRIPTION**

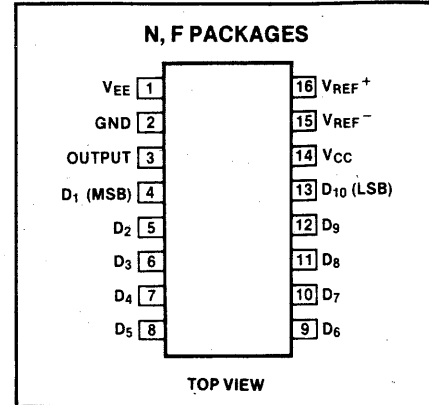
The NE5410/SE5410 are 10-Bit Multiplying Digital-to-Analog Converters pin-and-function compatible with the industry-standard MC3410, but with improved performance. These are capable of high-speed performance, and are used as general-purpose building blocks in cost-effective D/A systems.

The NE/SE5410 provides complete 10-bit accuracy and differential nonlinearity over temperature, and a wide compliance voltage range. Segmented current sources, in conjunction with an R/2R DAC provides the binary weighted currents. The output buffer amplifier and voltage reference have been omitted to allow greater speed, lower cost, and maximum user flexibility.

**FEATURES**

- Pin-and-function compatible with MC3410
- 10-bit resolution and accuracy ( $\pm 0.05\%$ )
- Guaranteed differential non-linearity over temperature
- Wide compliance voltage range— -2.5 to +2.5V
- Fast settling time—250ns typical
- Digital inputs are TTL and CMOS compatible
- High-speed multiplying input slew rate—20mA/ $\mu$ s
- Reference amplifier internally compensated
- Standard supply voltages +5V and -15V

**PIN CONFIGURATION**



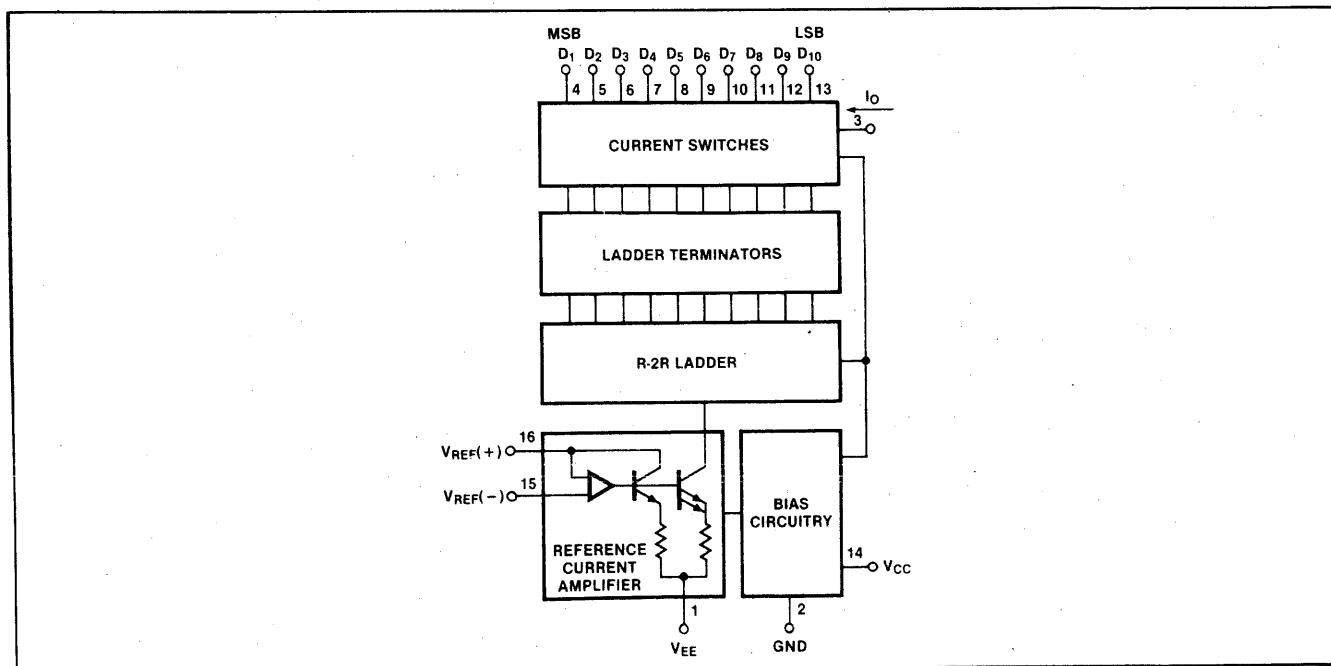
**APPLICATIONS**

- Successive approximation A/D converters
- High-speed, automatic test equipment
- High-speed modems
- Waveform generators
- CRT displays
- Strip CHART and X-Y plotters
- Programmable power supplies
- Programmable gain and attenuation

**ABSOLUTE MAXIMUM RATINGS**  $T_A = +25^\circ\text{C}$  unless otherwise noted

SYMBOL AND PARAMETER	RATING	UNIT
$V_{CC}$ Power Supply	+7.0	Vdc
$V_{EE}$	-18	Vdc
$V_I$ Digital Input Voltage	+15	Vdc
$V_O$ Applied Output Voltage	+4, -5.0	Vdc
$I_{REF(16)}$ Reference Current	2.5	mA
$V_{REF}$ Reference Amplifier Inputs	$V_{CC}, V_{EE}$	Vdc
$V_{REF(D)}$ Reference Amplifier Differential Inputs	0.7	Vdc
$T_A$ Operating Temperature Range		
SE5410	-55 to +125	$^\circ\text{C}$
NE5410	0 to +70	$^\circ\text{C}$
$T_J$ Junction Temperature		
Ceramic Package	+175	$^\circ\text{C}$
Plastic Package	+150	$^\circ\text{C}$

**BLOCK DIAGRAM**



LINEAR

Signetics

**SPATIAL, STEREO AND PSEUDO-STEREO SOUND CIRCUIT**

**TDA3810**

**Preliminary**

**DESCRIPTION**

The TDA3810 is an integrated circuit which can provide three switched functions for radio and television equipment: spatial sound from a stereo source; stereo sound from a stereo source; pseudo-stereo sound from a mono source.

**FEATURES**

- Three switched functions:  
 spatial  
 stereo  
 pseudo-stereo
- Muting circuit prevents LED flickering
- LED driving outputs (pins 7 and 8)
- TTL compatible inputs for selecting operating mode

**ABSOLUTE MAXIMUM RATINGS**

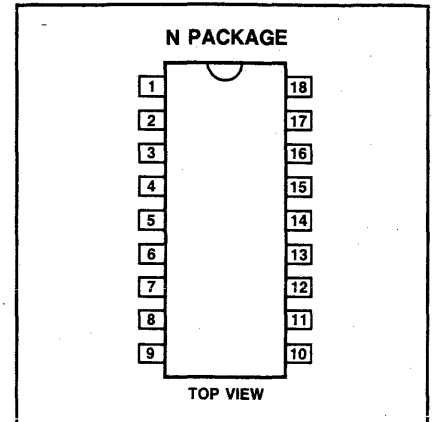
SYMBOL AND PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage (pin 18)	17	V
I <sub>CC</sub> Supply current (pin 18)	12	mA
T <sub>STG</sub> Storage temperature range	-25 to +150	°C
T <sub>A</sub> Operating ambient temperature range	0 to +70	°C
Thermal Resistance		
R <sub>th cr-a</sub> From crystal to ambient	80	K/W

**TRUTH TABLE**

MODE	CONTROL INPUT STATE		LED SPATIAL PIN 7	LED PSEUDO PIN 8
	PIN 11	PIN 12		
Mono pseudo-stereo	HIGH	LOW	Off	On
Spatial stereo	HIGH	HIGH	On	Off
Stereo	LOW	X	Off	Off

LOW = 0 to 0.8V (the less positive voltage)  
 HIGH = 2V to V<sub>CC</sub> (the more positive voltage)  
 X = state is don't care

**PIN CONFIGURATION**



**FUNCTIONAL PIN DESCRIPTION**

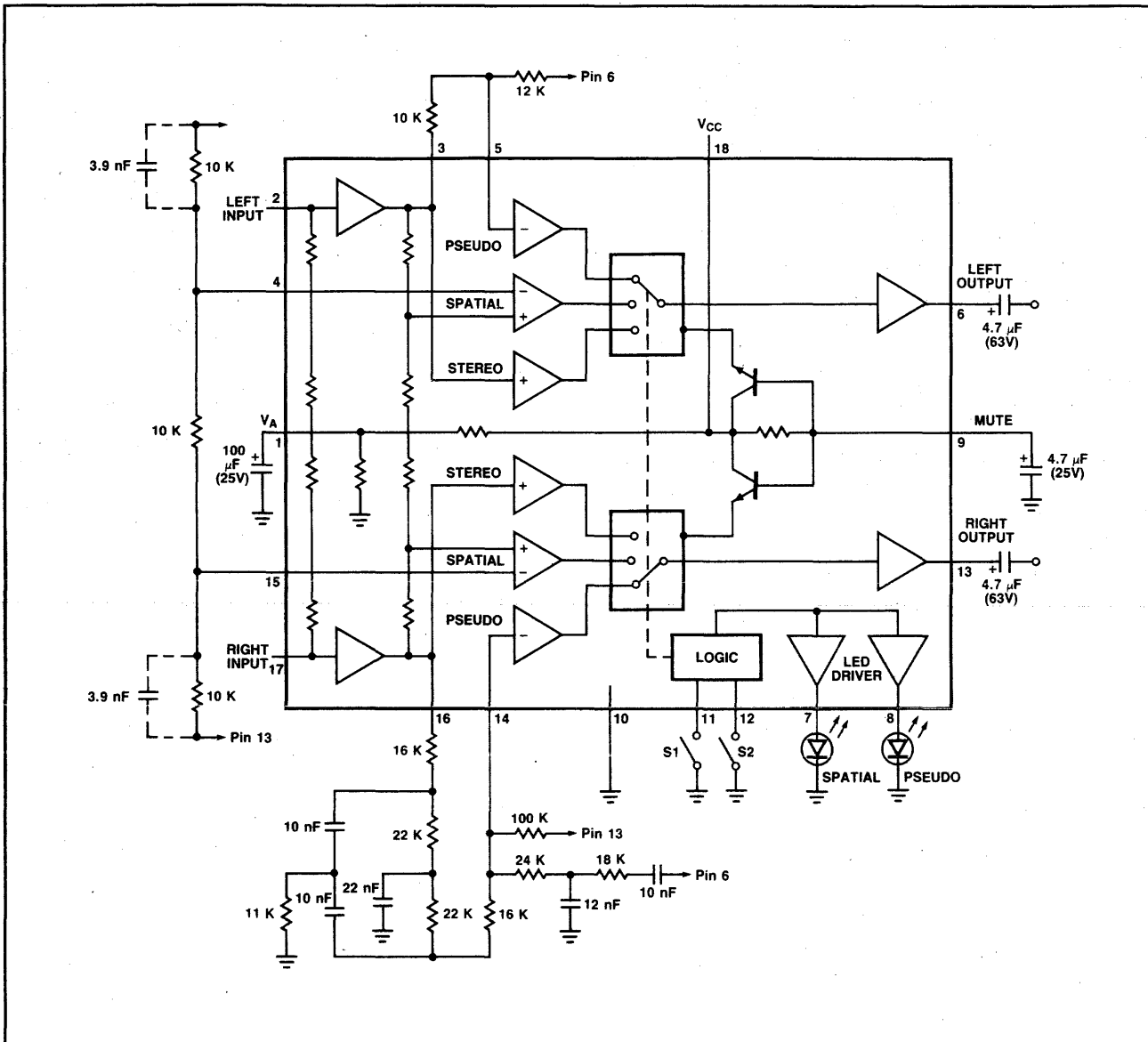
PIN NO.	NAME AND FUNCTION
1	V <sub>REF</sub>
2	Left channel input
3	Left channel stereo feedback
4	Left channel spatial feedback
5	Left channel pseudo feedback
6	Left channel output
7	Spatial indicator driver
8	Pseudo indicator driver
9	Mute
10	Ground
11	S1
12	S2
13	Right channel output
14	Right channel pseudo feedback
15	Right channel spatial feedback
16	Right channel stereo feedback
17	Right channel input
18	V <sub>DD</sub>

SPATIAL, STEREO AND PSEUDO-STEREO SOUND CIRCUIT

TDA3810

Preliminary

BLOCK DIAGRAM



LINEAR  
Signetics



**SPATIAL, STEREO AND PSEUDO-STEREO SOUND CIRCUIT**

**TDA3810**

**Preliminary**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 12V$ ;  $T_A = 25^\circ C$ ; Test circuit Figure 1 stereo mode (pin 11 to ground) unless otherwise specified.

SYMBOL AND PARAMETER	TEST CONDITION	TDA3810			UNIT
		Min	Typ	Max	
$V_{CC}$ Supply voltage range	(Pin 18)	4.5		16.5	V
$V_{CC}$ Supply current			6	12	mA
$V_S$ Reference voltage		5.3	6	6.7	V
$V_{i(rms)}$ Input voltage	(Pin 2 or 17) THD = 0.2%	2			V
$R_i$ Input resistance	(Pin 2 or 17)	50	75		k $\Omega$
$G_v$ Voltage gain $V_o/V_i$			0		dB
$\alpha$ Channel separation (R/L)		60	70		dB
THD Total harmonic distortion	$f = 40$ to $16000$ Hz; $V_{o(rms)} = 1V$		0.1		%
RR Power supply ripple rejection			50		dB
$V_{n(rms)}$ Noise output voltage	(Unweighted) left and right output		10		$\mu V$
Spatial mode (Pins 11 and 12 HIGH)					
$\alpha$ Antiphase crosstalk			50		%
$G_v$ Voltage gain		1.4	2.4	3.4	dB

**PSEUDO-STEREO MODE** The quality and strength of the pseudo-stereo effect is determined by external filter components.

SYMBOL AND PARAMETER	TEST CONDITION	TDA3810			UNIT
		Min	Typ	Max	
Control inputs	(Pins 11 and 12)				
$R_i$ Input resistance		70	120		k $\Omega$
$-I_i$ Switching current			35	100	$\mu A$
LED drivers (Pins 7 and 8)					
$I_o$ Output current for LED		10	12	15	mA
$V_F$ Forward voltage				6	V

COMPANDOR

NE570/571/SA571

DESCRIPTION

The NE570/571 is a versatile low cost dual gain control circuit in which either channel may be used as a dynamic range compressor or expander. Each channel has a full wave rectifier to detect the average value of the signal; a linearized, temperature compensated variable gain cell; and an operational amplifier.

The NE570/571 is well suited for use in telephone subscriber and trunk carrier systems, communications systems and hi-fi audio systems.

FEATURES

- Complete compressor and expander in 1 IC
- Temperature compensated
- Greater than 110dB dynamic range
- Operates down to 6Vdc
- System levels adjustable with external components
- Distortion may be trimmed out

CIRCUIT DESCRIPTION

The NE570/571 compandor building blocks, as shown in the block diagram, are a full wave rectifier, a variable gain cell, an operational amplifier and a bias system. The arrangement of these blocks in the IC result in a circuit which can perform well with few external components, yet can be adapted to many diverse applications.

The full wave rectifier rectifies the input current which flows from the rectifier input, to an internal summing node which is biased at V<sub>REF</sub>. The rectified current is averaged on an external filter capacitor tied to the CRECT terminal, and the average value of the input current controls the gain of the variable gain cell. The gain will thus be proportional to the average value of the input signal for capacitively coupled voltage inputs as shown in the following equation. Note that for capacitively coupled inputs there is no offset voltage capable of producing a gain error. The only error will come from the bias current of the rectifier (supplied internally) which is less than .1μA.

$$G \propto \frac{|V_{IN} - V_{REF}| \text{ avg.}}{R_1}$$

or

$$G \propto \frac{|V_{IN}| \text{ avg.}}{R_1}$$

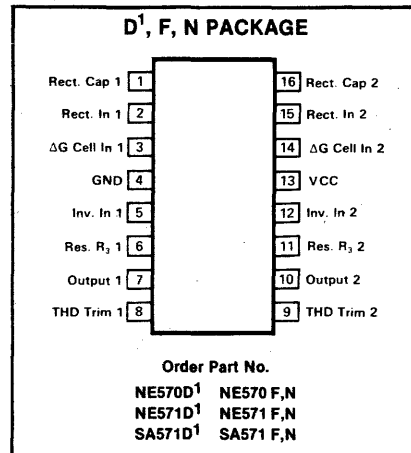
The speed with which gain changes to follow changes in input signal levels is determined by the rectifier filter capacitor. A small capacitor will yield rapid response but will not fully filter low frequency signals. Any ripple on the gain control signal will modulate the signal passing through the variable gain cell. In an expander or com-

Note:  
1. Supplied only in large SO (Small Outline) package.

APPLICATIONS

- Telephone trunk compandor—570
- Telephone subscriber compandor—571
- High level limiter
- Low level expander—noise gate
- Dynamic noise reduction systems
- Voltage controlled amplifier
- Dynamic filters

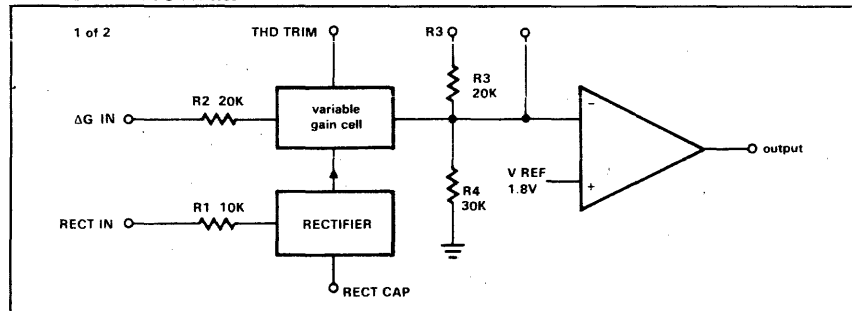
PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Positive supply	24	Vdc
TA Operating temperature range	0 to 70	°C
SA	-40 to +85	°C
PD Power dissipation	400	mW

BLOCK DIAGRAM



pressor application, this would lead to third harmonic distortion, so there is a tradeoff to be made between fast attack and decay times, and distortion. For step changes in amplitude, the change in gain with time is shown by this equation.

$$G(t) = (G_{\text{initial}} - G_{\text{final}}) e^{-t/\tau} + G_{\text{final}}$$

τ = 10K X C<sub>RECT</sub>

The variable gain cell is a current in, current out device with the ratio I<sub>OUT</sub>/I<sub>IN</sub> controlled by the rectifier. I<sub>IN</sub> is the current which flows from the ΔG input to an internal summing node biased at V<sub>REF</sub>. The following equation applies for capacitively coupled inputs. The output current, I<sub>OUT</sub>, is fed to the summing node of the op amp.

$$I_{IN} = \frac{V_{IN} - V_{REF}}{R_2} = \frac{V_{IN}}{R_2}$$

A compensation scheme built into the ΔG cell compensates for temperature, and cancels out odd harmonic distortion. The only distortion which remains is even harmonics, and they exist only because of internal offset voltages. The THD trim terminal provides a means for nulling the internal offsets for low distortion operation.

The operational amplifier (which is internally compensated) has the non-inverting input tied to V<sub>REF</sub>, and the inverting input connected to the ΔG cell output as well as brought out externally. A resistor, R<sub>3</sub>, is brought out from the summing node and allows compressor or expander gain to be determined only by internal components.

COMPANDOR

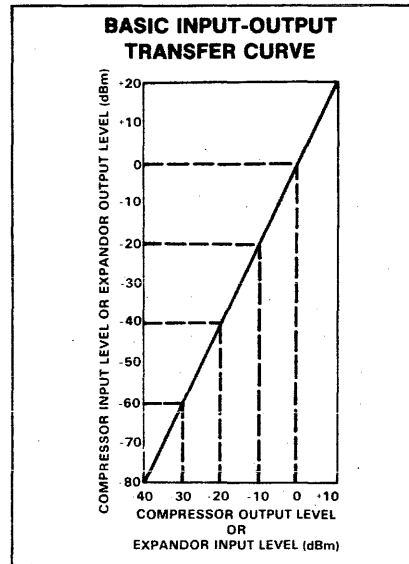
NE570/571/SA571

The output stage is capable of  $\pm 20\text{mA}$  output current. This allows a  $+13\text{dBm}$  ( $3.5\text{V rms}$ ) output into a  $300\Omega$  load which, with a series resistor and proper transformer, can result in  $+13\text{dBm}$  with a  $600\Omega$  output impedance.

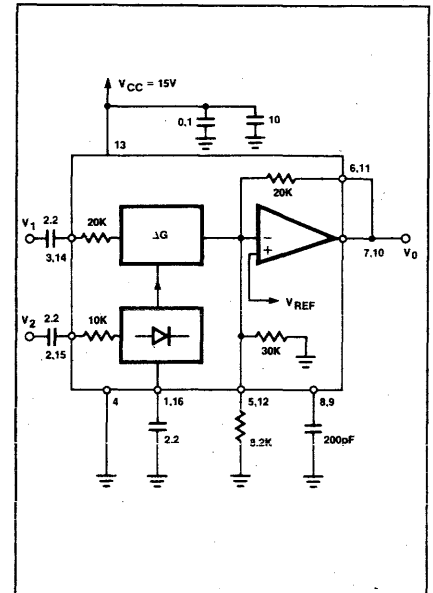
A band gap reference provides the reference voltage for all summing nodes, a regulated supply voltage for the rectifier and  $\Delta G$  cell, and a bias current for the  $\Delta G$  cell. The low tempo of this type of reference provides very stable biasing over a wide temperature range.

The typical performance characteristics illustration shows the basic input-output transfer curve for basic compressor or expander circuits.

TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL TEST CIRCUIT



DC ELECTRICAL CHARACTERISTICS  $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 15\text{V}$

PARAMETER	TEST CONDITIONS	NE570			NE/SA571 <sup>6</sup>			UNIT
		Min	Typ	Max	Min	Typ	Max	
$V_{CC}$ Supply voltage		6		24	6		18	V
$I_{CC}$ Supply current	No signal		3.2	4.0		3.2	4.8	mA
Output current capability		$\pm 20$			$\pm 20$			mA
Output slew rate			$\pm 5$			$\pm 5$		V/us
Gain cell distortion <sup>2</sup>	Untrimmed		.3	1.0		.5	2.0	%
	Trimmed		.05			.1		%
Resistor tolerance			$\pm 5$	$\pm 15$		$\pm 5$	$\pm 15$	%
Internal reference voltage		1.7	1.8	1.9	1.65	1.8	1.95	V
Output dc shift <sup>3</sup>	Untrimmed		$\pm 20$	$\pm 50$		$\pm 30$	$\pm 100$	mV
Expander output noise	No signal, 15Hz-20kHz <sup>1</sup>		20	45		20	60	$\mu\text{V}$
Unity gain level		-1	0	+1	-1.5	0	+1.5	dB
Gain change <sup>2,4</sup>	$-40^\circ\text{C} < T < 70^\circ\text{C}$		$\pm 1$			$\pm 1$		dB
	$0^\circ\text{C} < T < 70^\circ\text{C}$		$\pm 1$	$\pm 2$		$\pm 1$	$\pm 4$	dB
Reference drift <sup>4</sup>	$-40^\circ\text{C} < T < 70^\circ\text{C}$		+2, -25	-10, -40		+2, -25	+20, -50	mV
	$0^\circ\text{C} < T < 70^\circ\text{C}$		$\pm 5$	$\pm 10$		$\pm 5$	$\pm 20$	mV
Resistor drift <sup>4</sup>	$-40^\circ\text{C} < T < 70^\circ\text{C}$		+8, -0					%
	$0^\circ\text{C} < T < 70^\circ\text{C}$		+1, -0					%
Tracking error <sup>5</sup> , input $V_1 = \text{OdBm}$	Rectifier input, $V_2 =$ +6dBm -10dBm -20dBm -30dBm -40dBm		$\pm 2$ $\pm 2$ $\pm 2$ $\pm 2$ $\pm 2$	 -2,+4 -3,+6 -5,+1		$\pm 2$ $\pm 2$ $\pm 2$ $\pm 2$ $\pm 2$	 -2,+5 -4,+7 -1,+1.5	dB
Channel Separation		60			60			dB

NOTES

- Input to  $V_1$  and  $V_2$  grounded.
- Measured at  $\text{OdBm}$ , 1kHz
- Expander ac input change from no signal to  $\text{OdBm}$
- Relative to value at  $T_A = 25^\circ\text{C}$
- Relative to  $\text{OdBm}$
- Electrical characteristics for the SA571 only are specified over  $-40$  to  $+85^\circ\text{C}$  temperature range.

**PROGRAMMABLE ANALOG COMPANDOR**

**NE572**

**DESCRIPTION**

The NE572 is a dual channel, high performance gain control circuit in which either channel may be used for dynamic range compression or expansion. Each channel has a full wave rectifier to detect the average value of input signal; a linearized, temperature compensated variable gain cell ( $\Delta G$ ) and a dynamic time constant buffer. The buffer permits independent control of dynamic attack and recovery time with minimum external components and improved low frequency gain control ripple distortion over previous compandors.

The NE572 is intended for noise reduction in high performance audio systems. It can also be used in a wide range of communication systems and video recording applications.

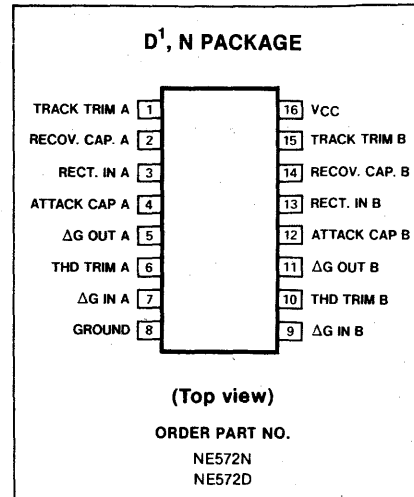
**FEATURES**

- Independent control of attack and recovery time.
- Improved low frequency gain control ripple
- Complementary gain compression and expansion with external Op Amp
- Wide dynamic range—greater than 110dB
- Temperature compensated gain control
- Low distortion gain cell
- Low noise— $6\mu V$  typical
- Wide supply voltage range—6V–22V
- System level adjustable with external components.

**APPLICATIONS**

- Dynamic noise reduction system
- Voltage control amplifier
- Stereo expander
- Automatic level control
- High level limiter
- Low level noise gate
- State variable filter

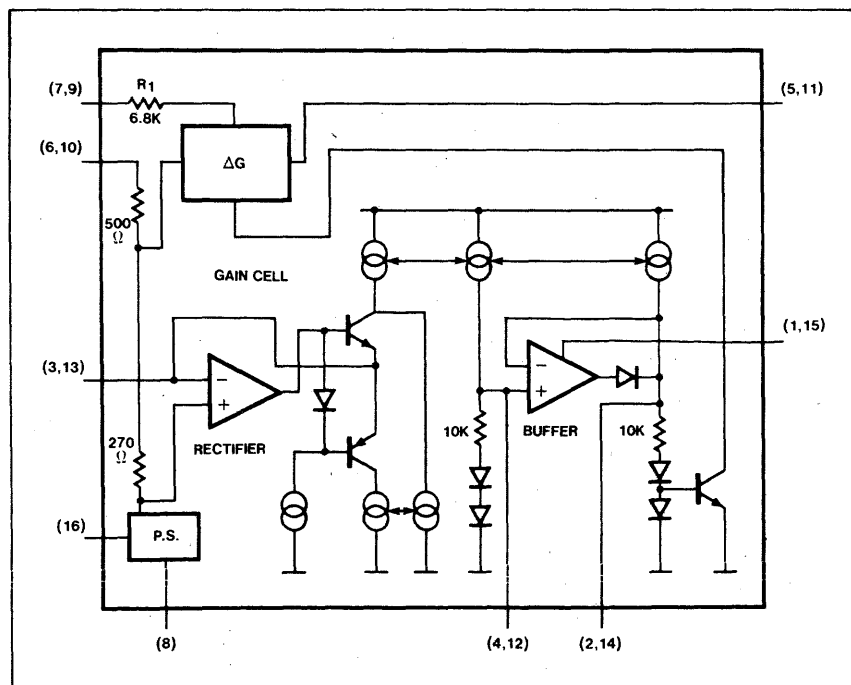
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT	
VCC	Supply voltage	22	VDC
T <sub>A</sub>	Operating temperature range	0 to 70	°C
P <sub>D</sub>	Power dissipation	500	mW

**BLOCK DIAGRAM**



Note:  
1. Supplied only in large SO (Small Outline) package.

LINEAR

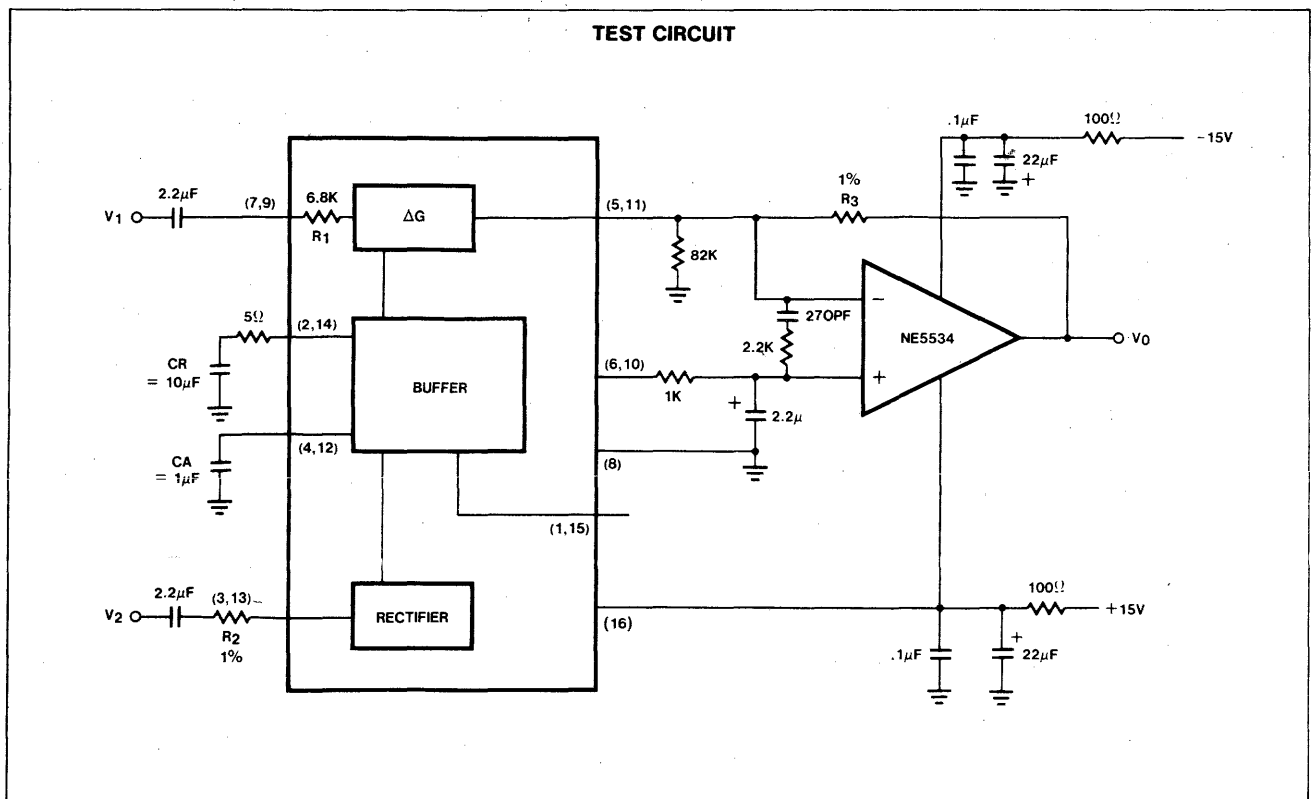
Signetics

**PROGRAMMABLE ANALOG COMPANDOR**

**NE572**

**ELECTRICAL CHARACTERISTICS** Standard Test Conditions (unless otherwise noted)  $V_{CC} = 15V$   $T_A = 25^{\circ}C$  Expander mode (see test circuit) Input signals at unity gain level = 100mV RMS at 1KHz,  $V_1 = V_2$ ,  $R_2 = 3.3K$   $R_3 = 17.3K$

PARAMETER	TEST CONDITIONS	LIMITS			UNIT
		Min	Typ	Max	
$V_{CC}$	Supply voltage	6		22	$V_{DC}$
$I_{CC}$	Supply current			6	mA
	Internal voltage reference	2.3	2.5	2.7	$V_{DC}$
THD	(untrimmed)		.2	1.0	%
THD	(trimmed)		.05		%
THD	(trimmed)		.25		%
	No signal output noise		6	25	$\mu V$
	DC level shift (untrimmed)		$\pm 20$	$\pm 50$	MV
	Unity gain level	-1	0	+1	dB
	Large signal distortion		0.7	3.0	%
	Tracking error measured relative to value at unity gain output		$\pm .2$	$\pm 1.5$	dB
	Channel crosstalk				dB
	Power supply rejection ratio		70		dB



**HIGH SPEED DUAL DIFFERENTIAL COMPARATOR/SENSE AMP**

**NE/SE521**

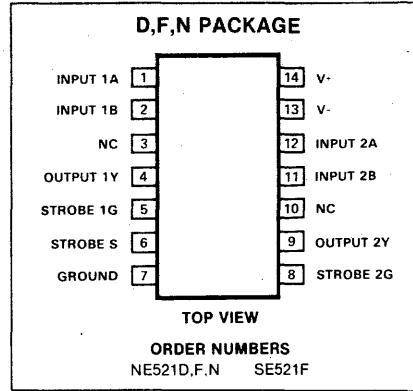
**FEATURES**

- 12ns maximum guaranteed propagation delay
- 20 $\mu$ A maximum input bias current
- TTL compatible strobes and outputs
- Large common mode input voltage range
- Operates from standard supply voltages
- Military qualifications pending

**APPLICATIONS**

- MOS memory sense amp
- A-to-D conversion
- High speed line receiver

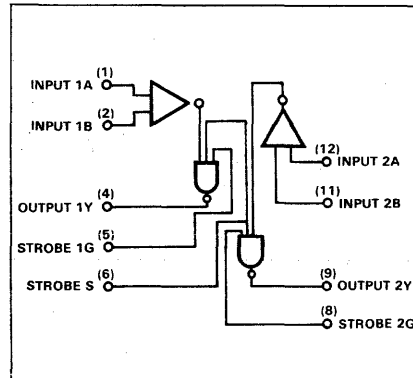
**PIN CONFIGURATION**



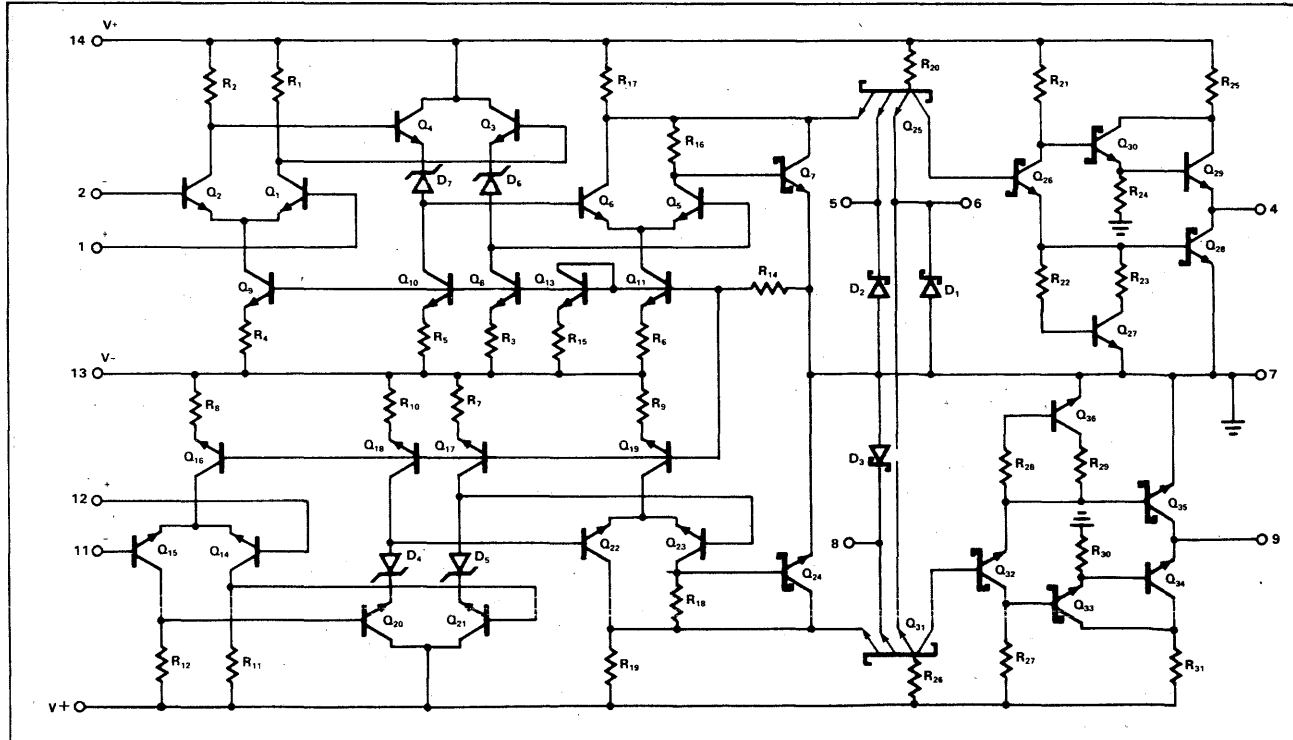
**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
Supply voltage		V
V+ Positive	+7	
V- Negative	-7	
V <sub>IDR</sub> Differential input voltage	$\pm 6$	V
V <sub>IN</sub> Input voltage		V
Common mode	$\pm 5$	
Strobe/gate	+5.25	
P <sub>D</sub> Power dissipation	600	mW
T <sub>A</sub> Operating temperature range		$^{\circ}$ C
NE521	0 to 70	
SE521	-55 to +125	
T <sub>stg</sub> Storage temperature range	-65 to +150	$^{\circ}$ C
Lead temperature (solder, 60 sec)	+300	$^{\circ}$ C

**BLOCK DIAGRAM**



**EQUIVALENT SCHEMATIC**



LINEAR

Signetics

**VOLTAGE COMPARATOR**

**NE/SE527**

**DESCRIPTION**

The SE/NE527 is a high speed analog voltage comparator which, in the first time mates state-of-the-art Schottky diode technology with the conventional linear process. This allows simultaneous fabrication of high speed T2L gates with a precision linear amplifier on a single monolithic chip. The SE/NE527 is similar in design to the Signetics SE/NE529 voltage comparator except that it incorporates a "Emitter Follower" input stage for extremely low input currents. This opens the door to a whole new range of applications for analog voltage comparators.

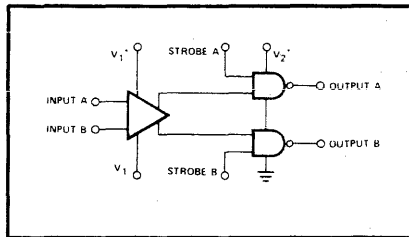
**FEATURES**

- 15ns propagation delay
- Complementary output gates
- TTL or ECL compatible outputs
- Wide common mode and differential voltage range
- Mil std 883A,B,C available
- Typical Gain of 5000

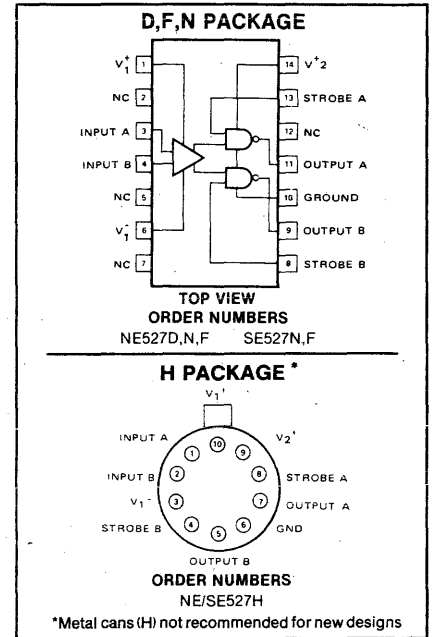
**APPLICATIONS**

- A/D conversion
- ECL to TTL interface
- TTL to ECL interface
- Memory sensing
- Optical data coupling

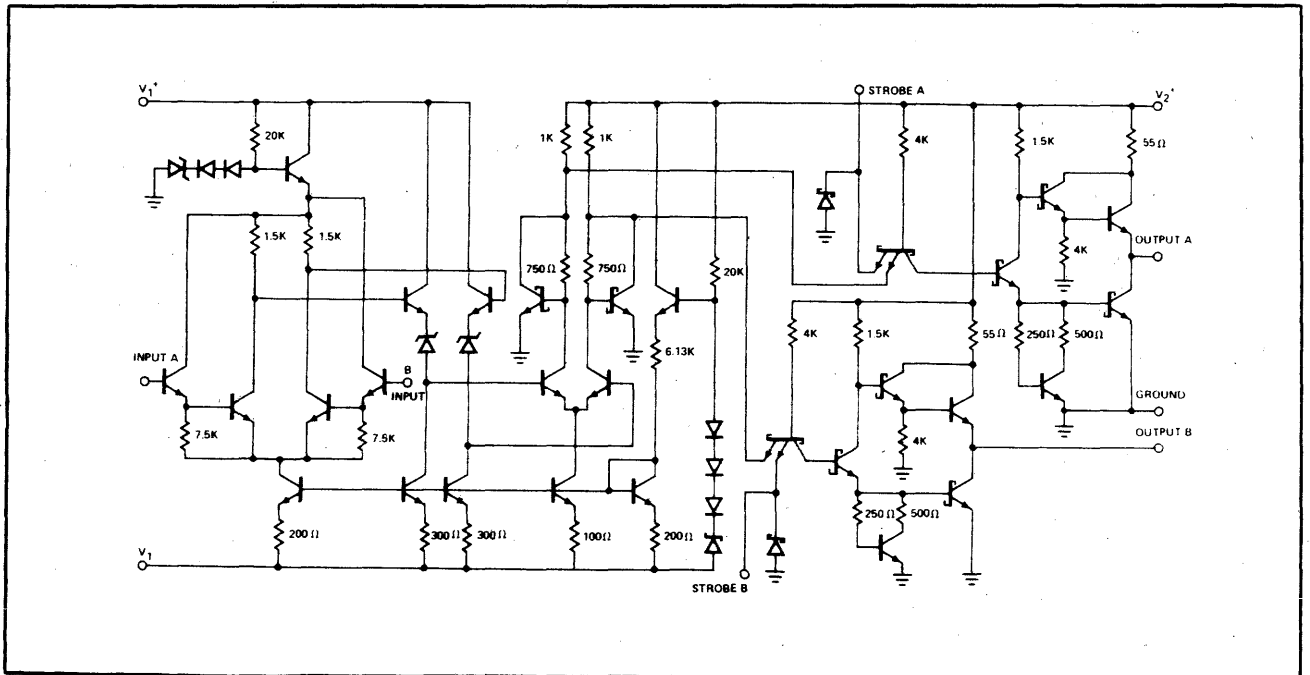
**BLOCK DIAGRAM**



**PIN CONFIGURATIONS**



**EQUIVALENT SCHEMATIC**



VOLTAGE COMPARATOR

NE/SE529

DESCRIPTION

The SE/NE529 is a high speed analog voltage comparator which, for the first time mates state-of-the-art Schottky diode technology with the conventional linear process. This allows simultaneous fabrication of high speed T<sup>2</sup>L gates with a precision linear amplifier on a single monolithic chip.

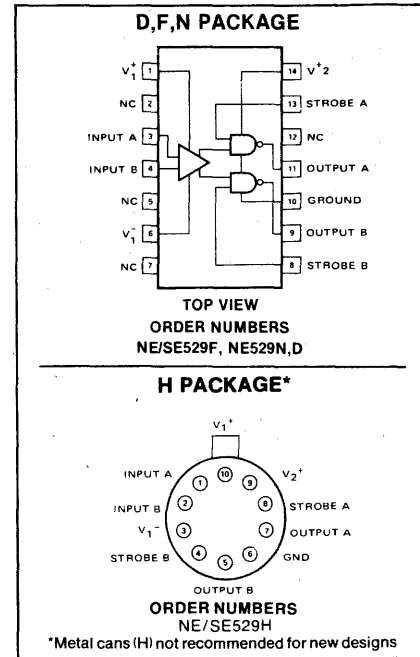
FEATURES

- 10ns propagation delay
- Complementary output gates
- TTL or ECL compatible outputs
- Wide common mode and differential voltage range
- Typical Gain 5000

APPLICATIONS

- A/D conversion
- ECL to TTL interface
- TTL to ECL interface
- Memory sensing
- Optical data coupling
- Mil std 883A,B,C available

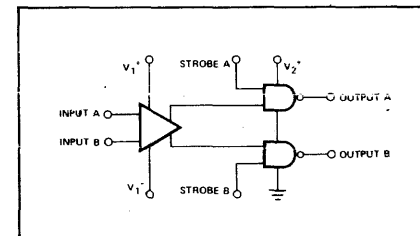
PIN CONFIGURATION



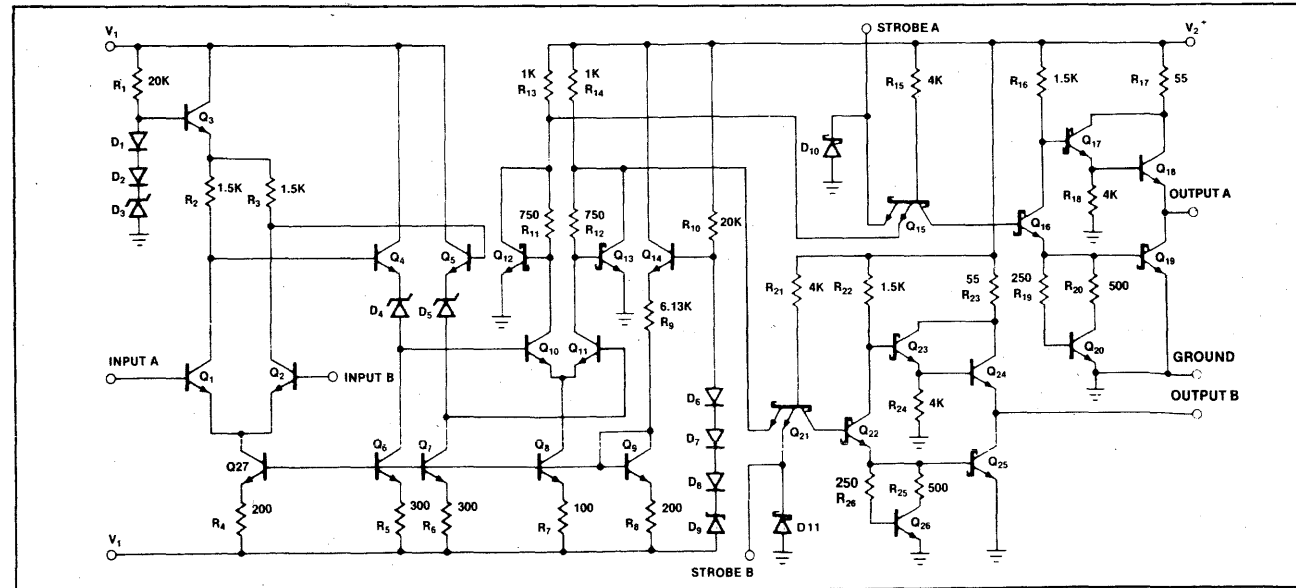
ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Positive supply voltage (V1+)	+15	V
Negative supply voltage (V1-)	-15	V
Gate supply voltage (V2+)	+7	V
Output voltage	+7	V
Differential input voltage	±5	V
Input common mode voltage	±6	V
Power dissipation	600	mW
Operating temperature range		
NE529	0 to +70	°C
SE529	-55 to +125	°C
Storage temperature range	-65 to +150	°C
Lead temperature (soldering, 60 sec)	+300	°C

BLOCK DIAGRAM



EQUIVALENT SCHEMATIC



LINEAR

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VACUUM FLUORESCENT DISPLAY DRIVER

NE/SA594

DESCRIPTION

The NE/SA594 is a display driver interface for vacuum fluorescent displays. The device is comprised of 8 drivers and a bias network and is capable of driving the digits and/or segments of most vacuum fluorescent displays.

The inputs are designed to be compatible with TTL, DTL, NMOS, PMOS or CMOS output circuitry.

There is an active pull-down circuit on each output so that display ghosting is minimized and no external components are required for most fluorescent display applications.

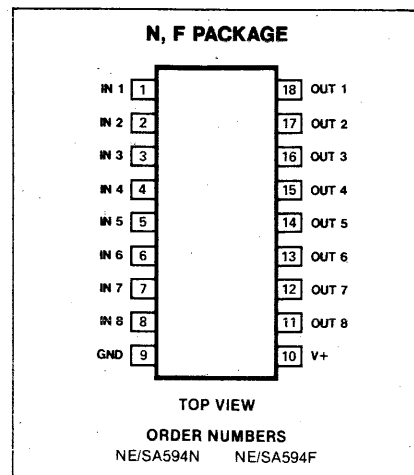
FEATURES

- Digit and/or segment drivers
- Active output pull-down circuitry
- High output breakdown voltage
- Low supply voltage
- Input compatible with all logic outputs

APPLICATIONS

- Digital clocks
- Dashboard displays
- Panel displays

PIN CONFIGURATION



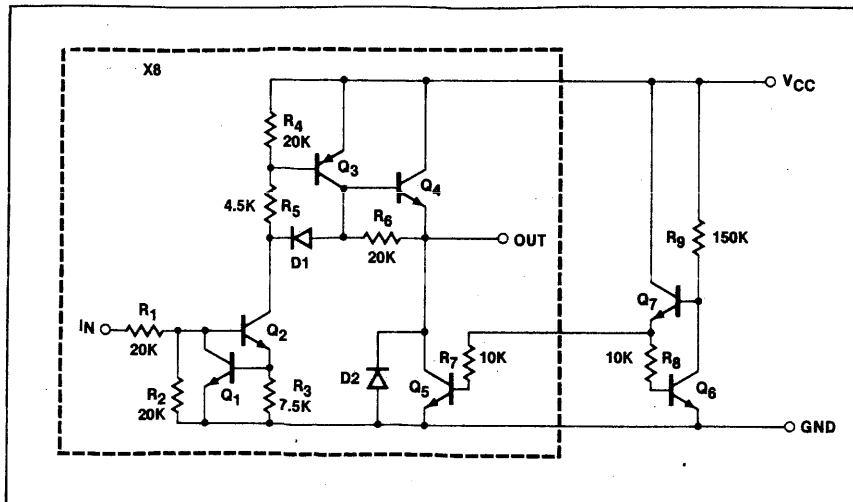
ABSOLUTE MAXIMUM RATINGS (at 25°C unless otherwise noted)

PARAMETER	RATING	UNIT
V <sub>CC</sub>	Supply voltage	45 V
V <sub>OUT</sub>	Output voltage	V <sub>CC</sub>
V <sub>IN</sub>	Input voltage	-0.3, +20 V
I <sub>OUT</sub>	Output current	
	Each output	50 mA
	All outputs	200 mA
P <sub>d</sub>	Power dissipation* (at 25°C)	800 mW
T <sub>A</sub>	Operating temperature range	
	NE	0 to 70 °C
	SA	-40 to +85 °C
T <sub>STG</sub>	Storage temperature range	-65 to +150 °C
T <sub>J</sub>	Maximum junction temperature	-165 °C
T <sub>SOLD</sub>	Lead soldering temperature (10 seconds)	300 °C

NOTE

\*Derate N (Plastic) Package above 38°C at 7.14 mW/°C.  
Derate F (Ceramic) Package above 75°C at 10.8 mW/°C.

EQUIVALENT SCHEMATIC



**ADDRESSABLE RELAY DRIVER**

**NE5090**

**DESCRIPTION**

The NE/SE5090 addressable relay driver is a high current latched driver, similar in function to the 9934 address decoder. The device has 8 open collector Darlington power outputs, each capable of 150mA load current. The outputs are turned on or off by respectively loading a logic "1" or logic "0" into the device data input. The required output is defined by a 3 bit address. The device must be enabled by a  $\overline{CE}$  input line which also serves the function of further address decoding. A common clear input,  $\overline{CLR}$ , turns all outputs off when a logic "0" is applied. The device is packaged in a 16 pin plastic or CERDIP package.

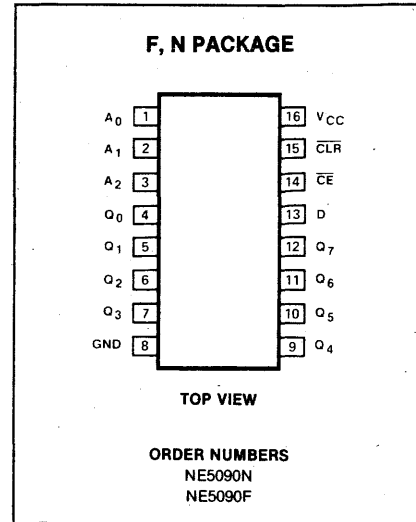
**FEATURES**

- 8 high current outputs
- Low-loading bus compatible inputs
- Power-on clear ensures safe operation
- Will operate in addressable or demultiplex mode
- Allows random (addressed) data entry
- Easily expandable
- Pin compatible with 9334

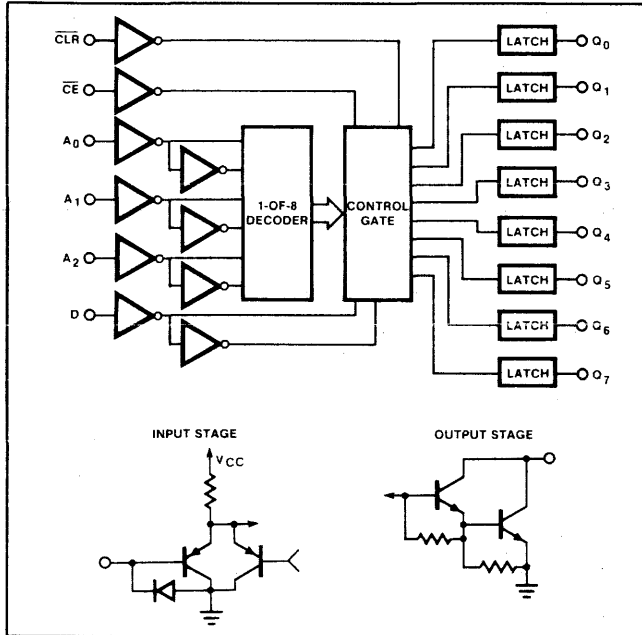
**APPLICATIONS**

- Relay driver
- Indicator lamp driver
- Triac trigger
- LED display digit driver
- Stepper motor driver

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**ABSOLUTE MAXIMUM RATINGS**

T<sub>A</sub> = 25°C unless otherwise specified.

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	- 0.5 to + 7	V
V <sub>IN</sub> Input voltage	- 0.5 to + 15	V
V <sub>OUT</sub> Output voltage	0 to + 30	V
I <sub>GND</sub> Ground current	500	mA
I <sub>OUT</sub> Output current	200	mA
Each output		
P <sub>D</sub> Power dissipation <sup>1</sup>	1	W
Ambient temperature range		
T <sub>A</sub> SE5090	- 55 to + 125	°C
T <sub>A</sub> NE5090	0 to + 70	
T <sub>J</sub> Junction	150	
T <sub>STG</sub> Storage	- 65 to + 150	
T <sub>sold</sub> Lead soldering temperature	300	°C
(10 sec max)		

LINEAR

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# ADDRESSABLE PERIPHERAL DRIVERS

# NE590/NE591

### DESCRIPTION

The NE590/591 addressable peripheral drivers are high current latched drivers, similar in function to the 9334 address decoder. The device has 8 Darlington power outputs, each capable of 250mA load current. The outputs are turned on or off by respectively loading a logic high or logic low into the device data input. The required output is defined by a 3-bit address. The device must be enabled by a  $\overline{CE}$  input line. A common clear input,  $\overline{CLR}$ , turns all outputs off when a logic low is applied.

The NE590 has 8 open collector Darlington outputs which sink current to ground. The device is packaged in a 16-pin molded or cerdip package.

The NE591 has 8 open emitter Darlington outputs which source current to an external load from a common collector line,  $V_S$ . This  $V_S$  line need not necessarily be the same as the 5 volt  $V_{CC}$  supply. The device is packaged in an 18-pin molded or cerdip package.

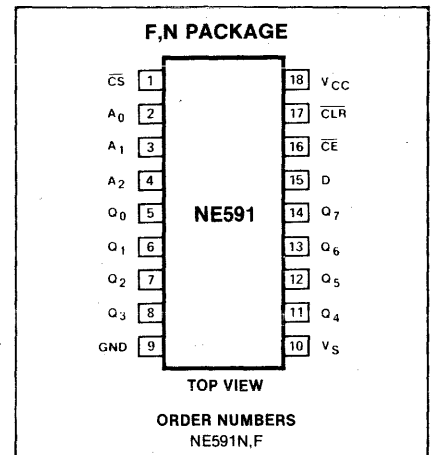
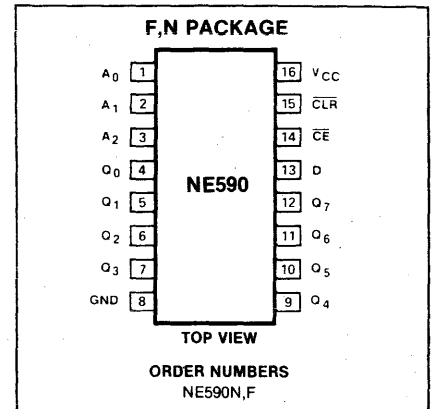
### FEATURES

- 8 high current outputs
- Low-loading bus compatible inputs
- Power-on clear ensures safe operation
- NE590 will operate in addressable or demultiplex mode
- Allows random (addressed) data entry
- Easily expandable
- NE590 is pin compatible with 9334

### APPLICATIONS

- Relay driver
- Indicator lamp driver
- Triac trigger
- LED display digit driver
- Stepper motor driver

### PIN CONFIGURATION



### PIN DESIGNATION

590 PIN NO.	591 PIN NO.	SYMBOL	NAME & FUNCTION
1-3	2-4	A <sub>0</sub> -A <sub>2</sub>	A 3-bit binary address on these pins defines which of the 8 output latches is to receive the data.
4-7, 9-12	5-8, 11-14	Q <sub>0</sub> -Q <sub>7</sub>	The 8 device outputs. The NE590 has open collector Darlington outputs. The NE591 has open emitter follower outputs.
13	15	D	The data input. When the chip is enabled, this data bit is transferred to the defined output such that: "1" turns output switch "ON" "0" turns output switch "OFF"
14	16	$\overline{CE}$	Thus in logic terms, the NE590 inverts data to the relevant output. The NE591 retains true data at the output. The chip enable. When this input is low, the output latches will accept data. When $\overline{CE}$ goes high, all outputs will retain their existing state, regardless of address or data input conditions.
15	17	$\overline{CLR}$	The clear input. When $\overline{CLR}$ goes low all output switches are turned "OFF". On the NE590, a high data input will override the clear function on the addressed latch. On the NE591, $\overline{CLR}$ low will override any other condition.
—	1	$\overline{CS}$	The chip select input provides for an additional level of address decoding.
—	10	$V_S$	The $V_S$ line provides the power to all 8 output devices. It is connected to the collectors of all 8 output transistors. This pin may be connected to the $V_{CC}$ or another supply.

**PROGRAMMABLE SEVEN CHANNEL RC ENCODER**

**NE5044**

**DESCRIPTION**

The NE5044 is a programmable parallel input, serial output pulsewidth encoder. A multiplexed dual linear ramp technique is used to allow up to 7 inputs to be converted to a serial pulsewidth modulated signal with excellent linearity and minimal crosstalk. Fixed or variable frame rates can be used, externally controlled, for ease of demodulation. An onboard 5V regulator eliminates power supply sensitivities and provides up to 20mA current capability for driving external loads.

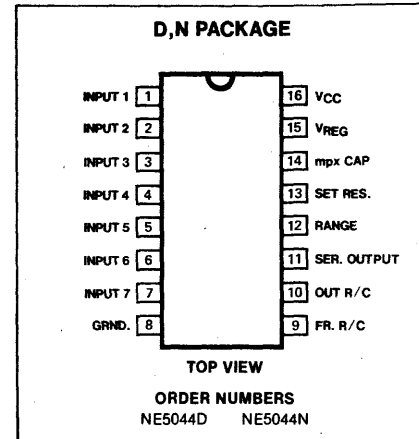
**FEATURES**

- 3 to 7 channels, externally selectable
- Constant current dual linear ramp for linearity better than .3%
- Internal voltage regulator for low drift
- Wide supply range 4.5 – 16V
- Fixed or variable frame rate set by external R-C
- External control for channel gain or range
- Versatile applications; exponential rates, mixing, dual rate, reversing etc.
- Compatible with all transmission mediums

**APPLICATIONS**

- Radio controlled aircraft, cars, boats, trains
- Industrial controllers
- Remote controlled entertainment systems
- Security systems
- Instrumentation recorders/controls
- Remote Analog/digital data transmission
- Automotive sensor systems
- Robotics
- Telemetry

**PIN CONFIGURATION**



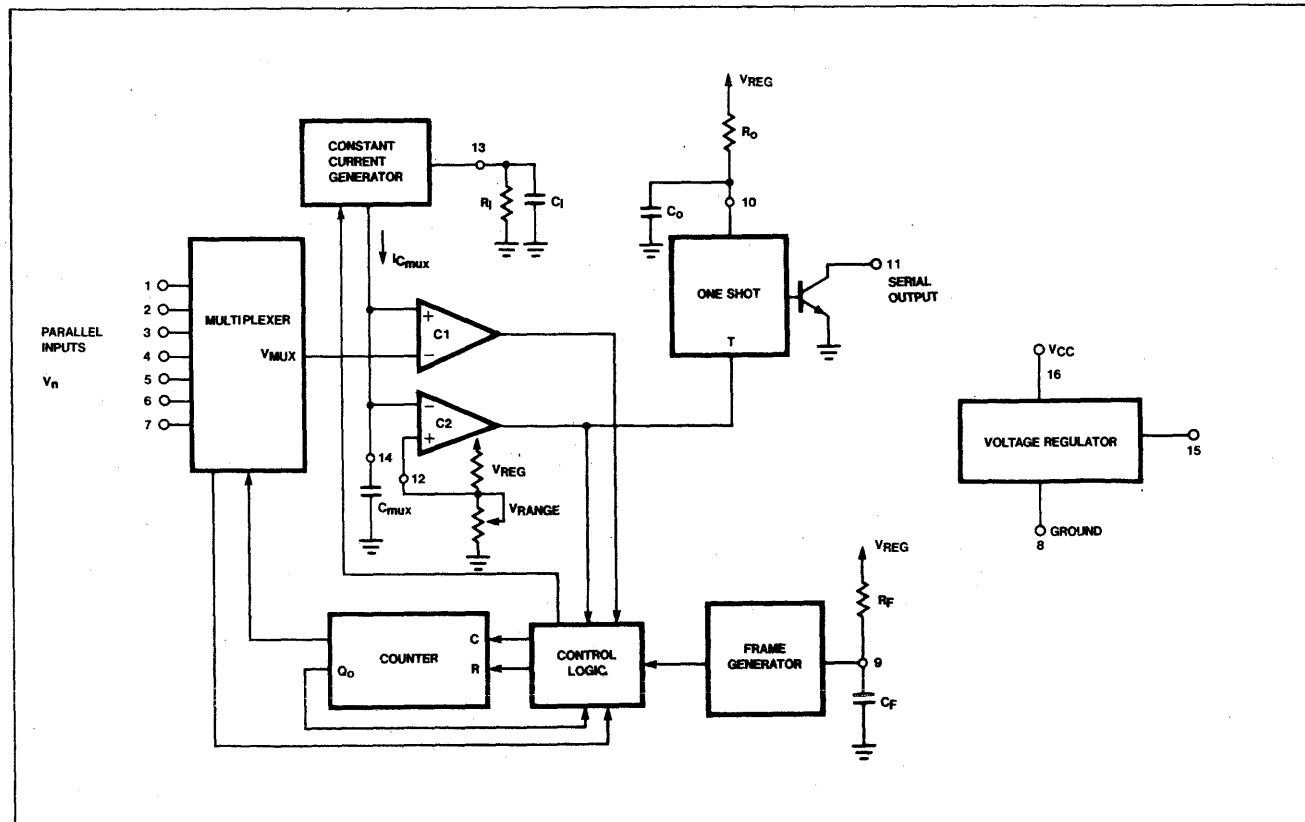
**ABSOLUTE MAXIMUM RATINGS<sup>1</sup>**

PARAMETER	RATING	UNIT
V <sub>CC</sub> , Supply voltage	17	V
Regulator output current	-25	mA
Serial output peak current	30	mA
Constant current generator	-1	mA
Parallel inputs, range input	0-V <sub>REG</sub>	V
One shot input, frame generator input	0-V <sub>REG</sub>	V
Operating temperature	-20 to +75	°C
Storage temperature	-65 to +150	°C

NOTE

1. T<sub>A</sub> = 25° unless otherwise stated.

**BLOCK DIAGRAM**



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PROGRAMMABLE SEVEN CHANNEL RC ENCODER

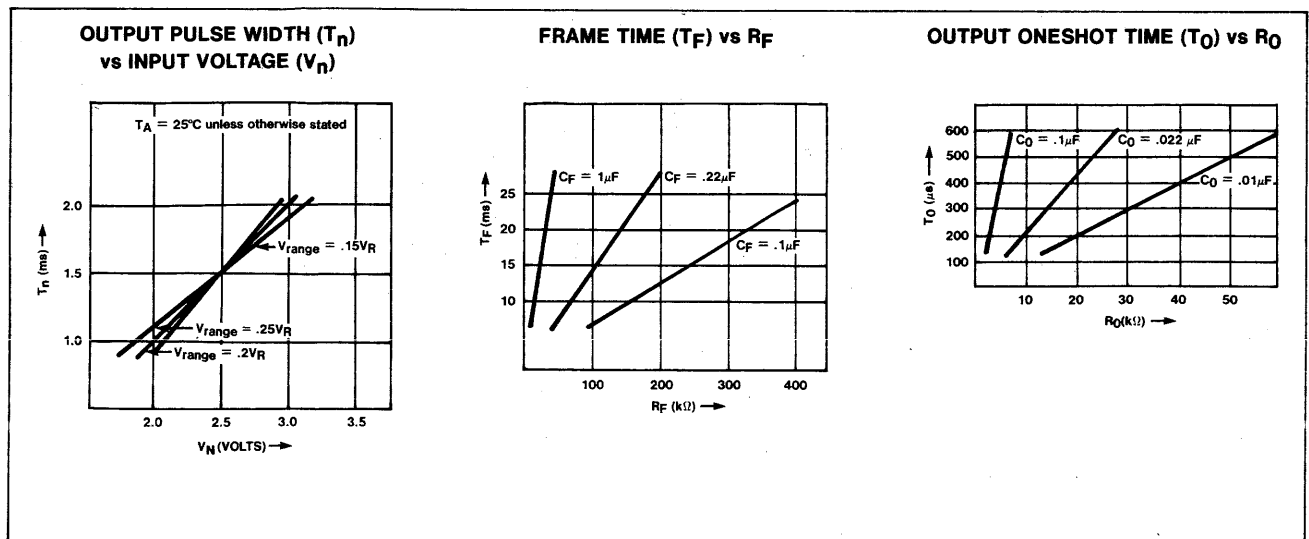
NE5044

DC ELECTRICAL CHARACTERISTICS Test conditions  $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 10\text{V}$  using Test Circuit A unless otherwise stated.

PARAMETER	TEST CONDITIONS	NE 5044			UNIT	
		Min	Typ	Max		
POWER SUPPLY REQUIREMENTS (Note 1) Power supply voltage range Power supply current	Excluding control pots and serial output currents	4.5		16	V	
			11	15	mA	
$V_{REG}$ VOLTAGE REGULATOR Output voltage Output current Line regulation	$V_R \geq 4.5\text{V}$ $7 \leq V_{CC} \leq 16$	4.5	5.0	5.5	V	
				-20	mA	
			.005	0.2	V/V	
MULTIPLEXER Input current Input voltage range Crosstalk	$V_n = 2.5\text{v}$ $V_n - V_{Range} \geq .75\text{V}$		$\pm 30$	$\pm 200$	nA	
		1.5		5	V	
			$\pm 1$	$\pm 5$	$\mu\text{s}$	
$T_n$ OUTPUT PULSE Position Position linearity error Position tempco Position PSR	$R_I \cdot C_{mux} = 1.25\text{ms}$ $V_n = .5V_{REG}; V_{RANGE} = .2V_{REG}$ $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $6\text{V} \leq V_{CC} \leq 16\text{V}$	1350	1500	1650	$\mu\text{s}$	
				5	$\mu\text{s}$	
				.15	$\mu\text{s}/^\circ\text{C}$	
				.5	$\mu\text{s}/\text{V}$	
				1	$\mu\text{s}/\text{V}$	
$T_0$ Width Saturation voltage Leakage current Range input voltage Frame time (Fixed) Inhibit threshold	$R_0C_0 = 300\mu\text{s}$ $I_0 = 25\text{mA}$ $R_I = 50\text{k}\Omega$ $R_I = 25\text{k}\Omega$ $R_F C_F = 30\text{ms}$	240	285	330	$\mu\text{s}$	
				.6	1	V
				.05	50	$\mu\text{A}$
			.75			V
			1.00			V
			17	20	23	ms
					.4	V

NOTE

1. At supply voltages exceeding 12V, a current limiting resistor of 20 to 50 $\Omega$  in series with  $V_{CC}$  is recommended.

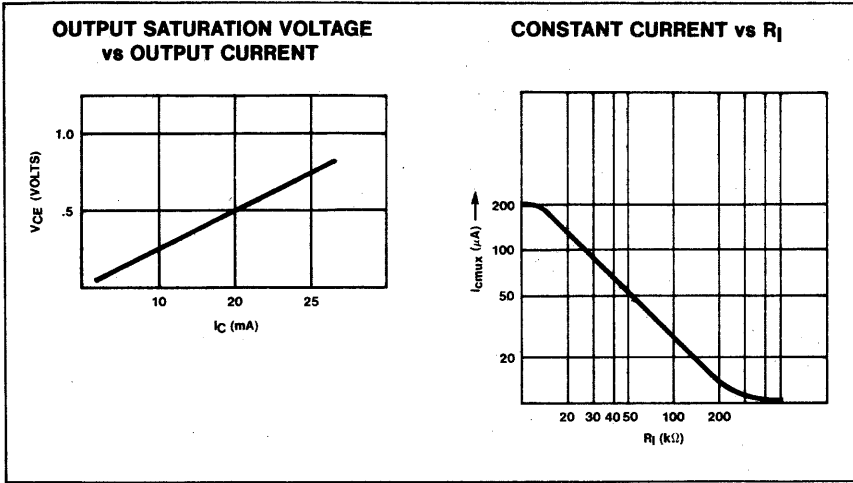


LINEAR

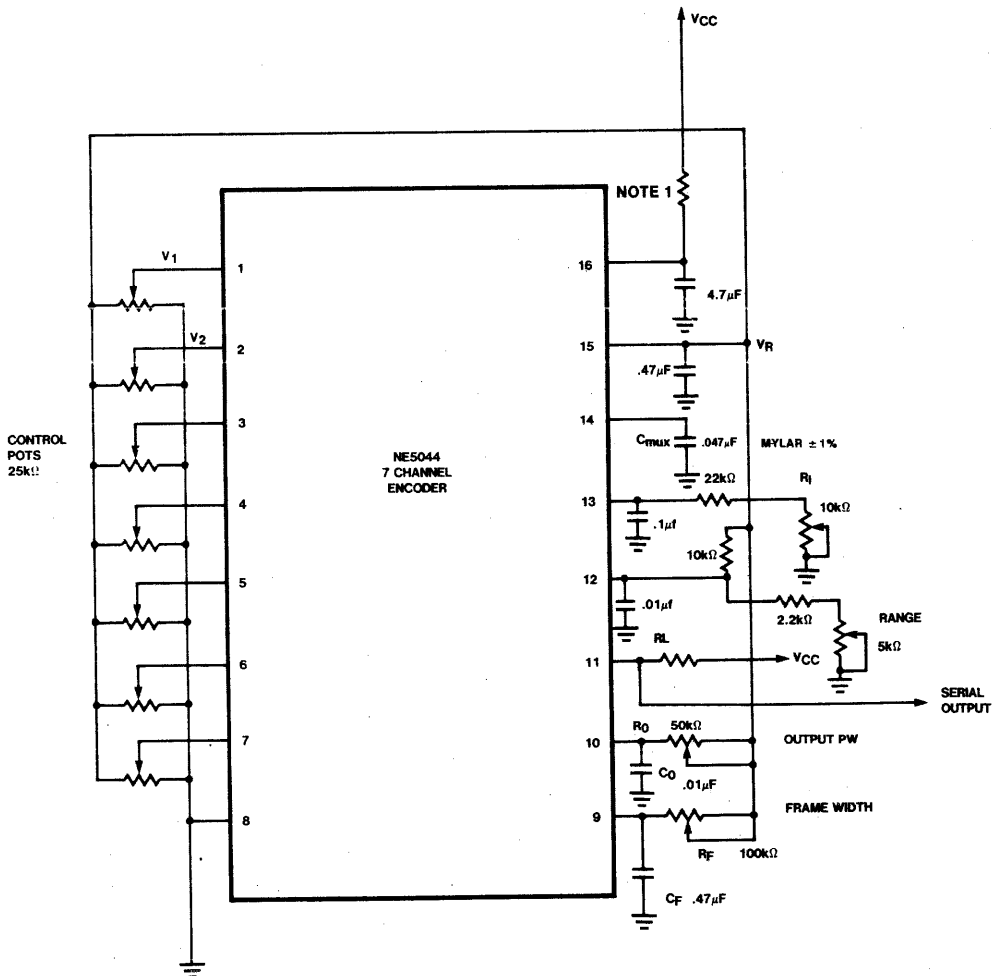
Signetics

**PROGRAMMABLE SEVEN CHANNEL RC ENCODER**

**NE5044**



**TEST CIRCUIT**



NOTE  
1. At supply voltages exceeding 12V, a current limiting resistor of 20 to 50Ω in series with VCC is recommended

LINEAR

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**SEVEN CHANNEL RC DECODER**

**NE5045**

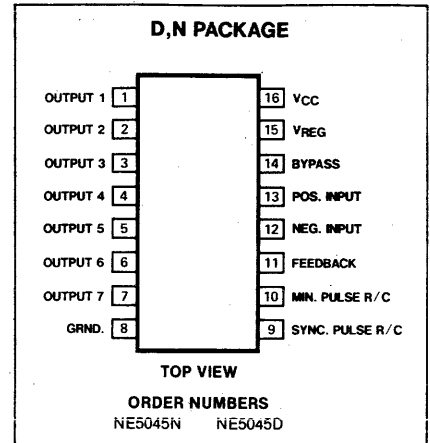
**DESCRIPTION**

The NE5045 is a serial input, parallel output, decoder intended for applications in pulse width or pulse position modulation systems. The serial input pulse, either positive or negative, is shaped and amplified before being fed to the counter/decoder. An integrating type sync. separator detects pulses greater than  $T_w = R_s C_s$ . The amplified input pulse triggers an internal one-shot (minimum pulse) which in turn clocks the counter-decoder, thereby enhancing system noise rejection. A missing pulse detector resets the decoder during the sync. pause. An internal voltage regulator supplies power for the radio receiver providing excellent isolation from the power supply as well as the decoder logic.

**APPLICATIONS**

- Radio controlled aircraft, cars, boats, trains
- Industrial controllers
- Remote controlled entertainment systems
- Security systems
- Instrumentation recorders/controls
- Remote Analog/digital data transmission
- Automotive sensor systems
- Robotics
- Telemetry

**PIN CONFIGURATION**



**FEATURES**

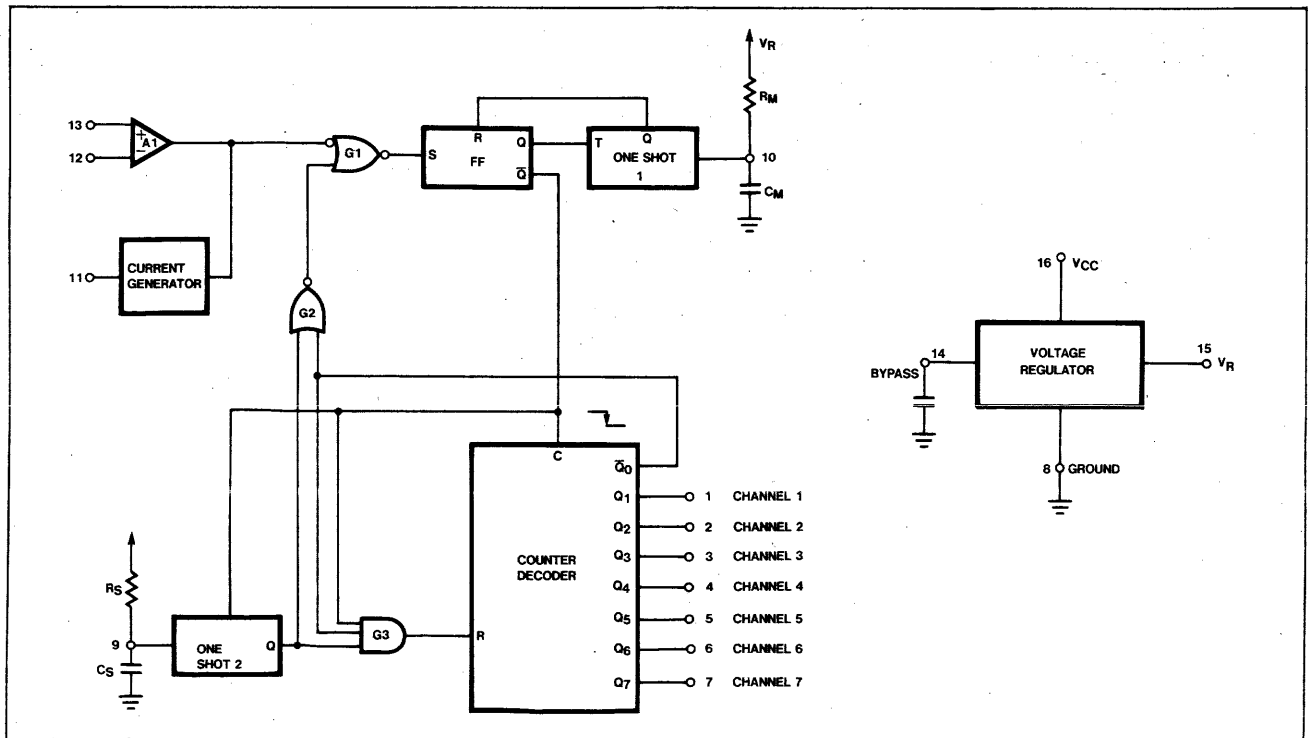
- Decodes up to 7 channels
- High gain input amplifier
- Externally set sync. pause and minimum pulse
- Wide supply voltage range, 3.6V-8V.
- Positive or negative pulse inputs
- Noise and flutter rejection
- Outputs reset to zero without inputs
- Compatible with all transmission mediums

**ABSOLUTE MAXIMUM RATINGS<sup>1</sup>**

PARAMETER	RATING	UNIT
V <sub>CC</sub> , Supply voltage	10	V
Regulator output current	-25	mA
Decoded output current	±5	mA
Pause input voltage	0 to V <sub>R</sub>	V
Input amplifier voltage	0 to V <sub>R</sub>	V
Operating temperature	-20 to +75	°C
Storage temperature	-65 to +150	°C

NOTE  
1. T<sub>A</sub> = 25°C unless otherwise stated

**BLOCK DIAGRAM**



**SEVEN CHANNEL RC DECODER**

**NE5045**

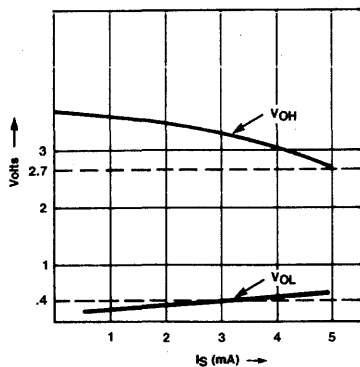
**DC ELECTRICAL CHARACTERISTICS** Standard conditions: ( $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 5.0\text{V}$  unless otherwise stated), using Test Circuit # 1

PARAMETER	TEST CONDITIONS	NE5045			UNIT
		Min	Typ	Max	
<b>POWER SUPPLY REQUIREMENTS</b> Power supply voltage range Power supply current	Test circuit # 1 Excluding input bias current	3.6	9.0	8.0 14.0	V mA
<b>V<sub>R</sub></b> VOLTAGE REGULATOR Output voltage Output current Line regulation Voltage drop	$V_R \geq 3.7\text{V}$ $V_{CC} = 6\text{V to } 8\text{V}$ $V_{CC} = 4\text{V}, I_R = -10\text{mA}$	3.7	4.1 .01	4.5 -15 .05 1.3	V mA V/V V
<b>T<sub>S</sub></b> <b>T<sub>M</sub></b> INPUT AMPLIFIER Input bias current Input voltage range Open loop gain Feedback current Detection threshold Sync. pause time Minimum pulse time	Test circuit # 1, $\Delta V_{12} \& 13$ $R_S C_S = 6.0\text{ms}$ $R_M C_M = 500\mu\text{s}$	2.0 100 5.1 405	10 60 200 8 6.0 475	100 4.0 400 20 6.9 545	nA V dB $\mu\text{A}$ mV ms $\mu\text{s}$
<b>OUTPUTS-ALL CHANNELS</b> V <sub>OL</sub> V <sub>OH</sub>	$I_{\text{SINK}} = 1\text{mA}$ $I_{\text{SOURCE}} = 2\text{mA}$	2.7	.25	.5	V V

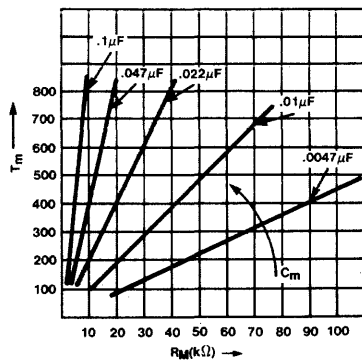
LINEAR

Signetics

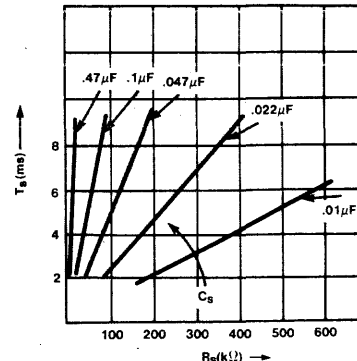
**V<sub>OL</sub> vs SINK CURRENT AND V<sub>OH</sub> vs SOURCE CURRENT**



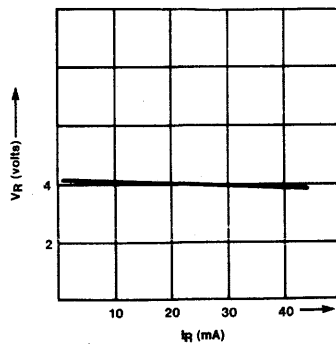
**MINIMUM PULSE TIME, T<sub>M</sub> vs R<sub>M</sub>, C<sub>M</sub>**



**SYNC. PAUSETIME, T<sub>S</sub> vs R<sub>S</sub>C<sub>S</sub>**



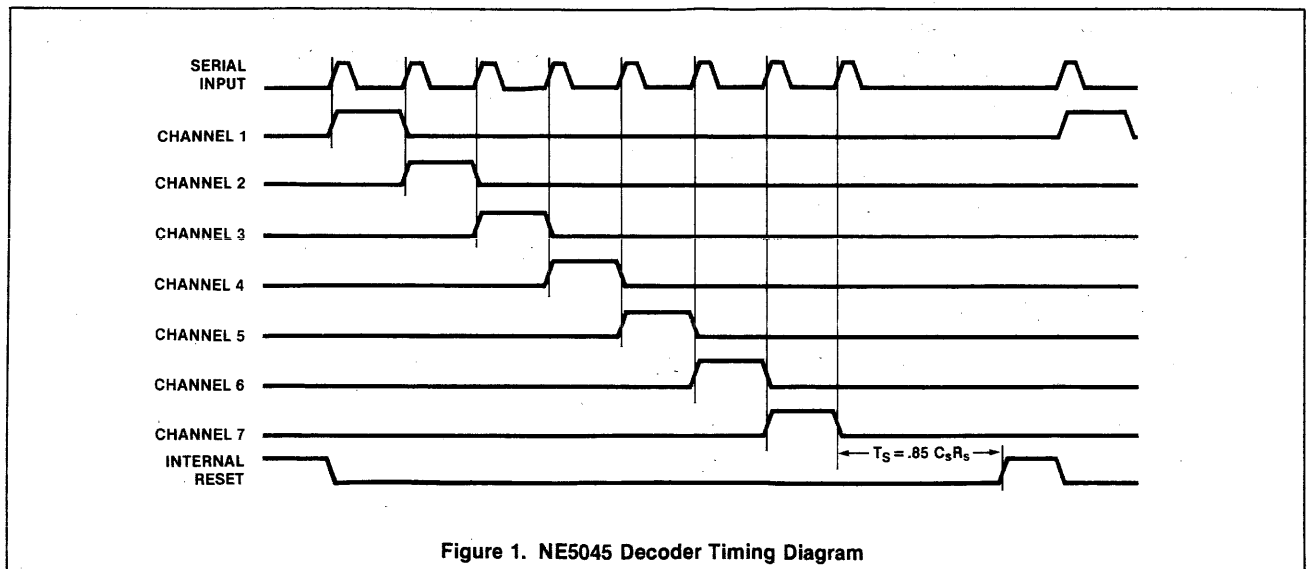
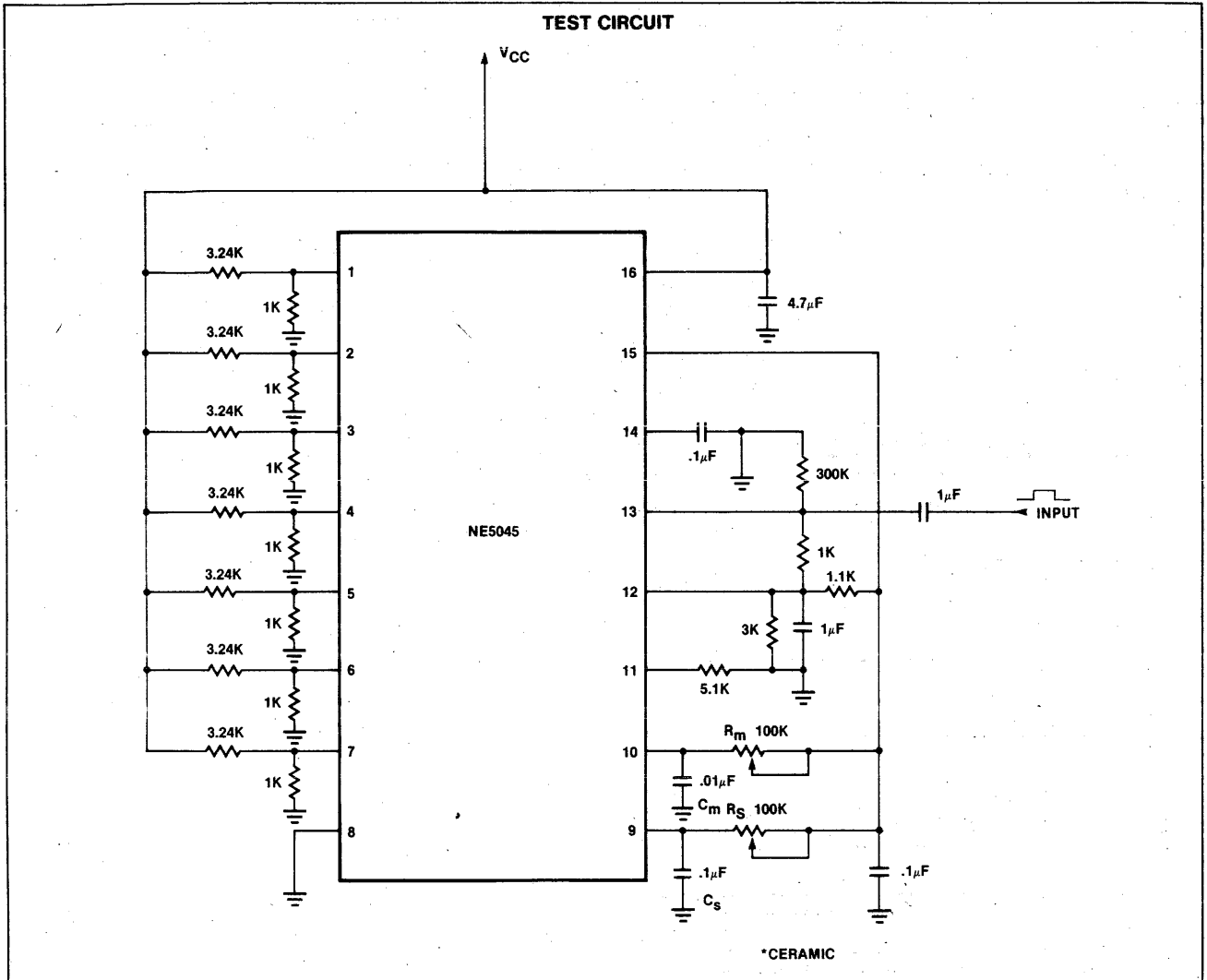
**REGULATOR VOLTAGE vs LOAD CURRENT**





SEVEN CHANNEL RC DECODER

NE5045



LINEAR  
Signetics

**SERVO AMPLIFIER**

**NE544**

**DESCRIPTION**

The NE544 is a servo amplifier and pulse-width demodulator with internal motor drive transistors. It is designed for remote control applications in digital proportional systems but can be used in many other closed loop position control applications. It incorporates a linear one shot for improved positional accuracy and outputs for external pnp motor drive transistors.

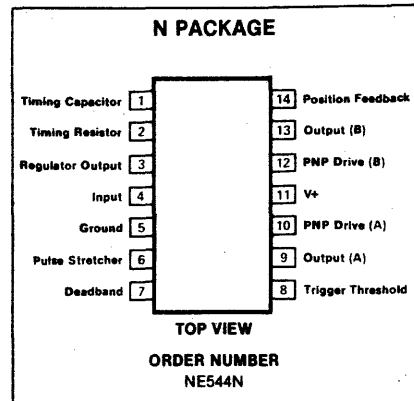
**FEATURES**

- 500mA load current capability
- Bidirectional bridge output with single power supply
- Low standby power drain
- Adjustable deadband and trigger thresholds
- High linearity, 0.5% maximum error
- Output drive for external PNP transistors (optional)
- Wide supply voltage range

**APPLICATIONS**

- Miniature position Servo
- Robotics
- Control devices
- Remote positioning

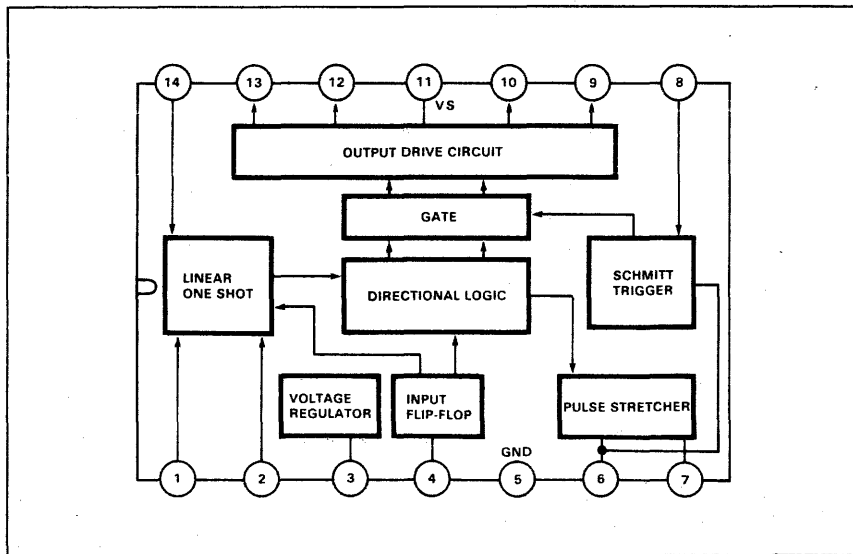
**PIN CONFIGURATIONS**



**ABSOLUTE MAXIMUM RATINGS**  $T_A = 25^\circ\text{C}$  unless otherwise specified.

PARAMETER	RATING	UNIT
V+	Supply voltage	6.0 V
$I_O$	Output current	500 mA
$T_A$	Operating temperature	-20 to +75 °C
$T_{stg}$	Storage temperature	-65 to +150 °C

**BLOCK DIAGRAM**



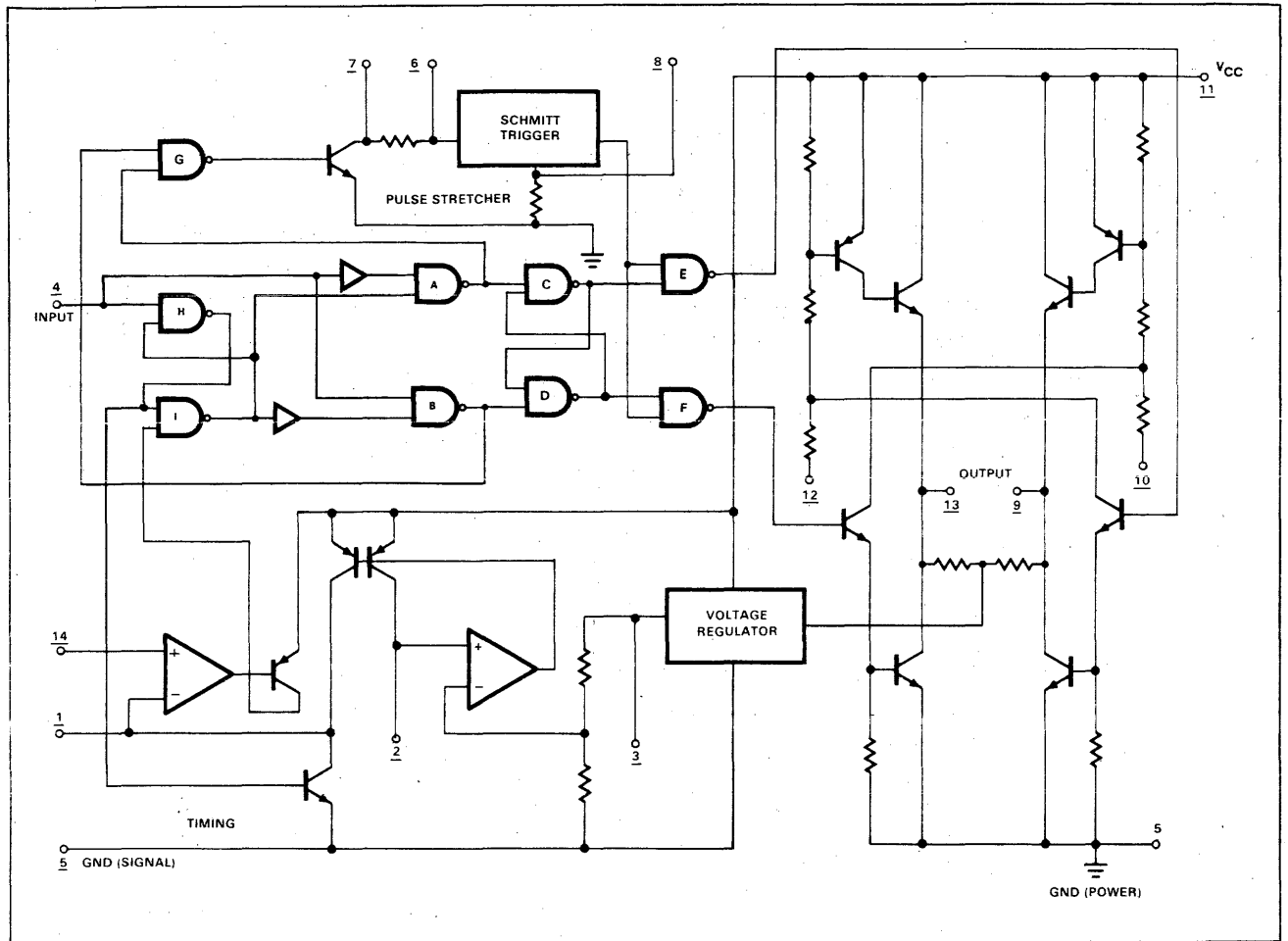
LINEAR

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**SERVO AMPLIFIER**

**NE544**

**EQUIVALENT CIRCUIT SCHEMATIC**



**DC ELECTRICAL CHARACTERISTICS**  $T_A = 25^\circ\text{C}$ ,  $V_S = 4.8\text{V}$  unless otherwise specified.

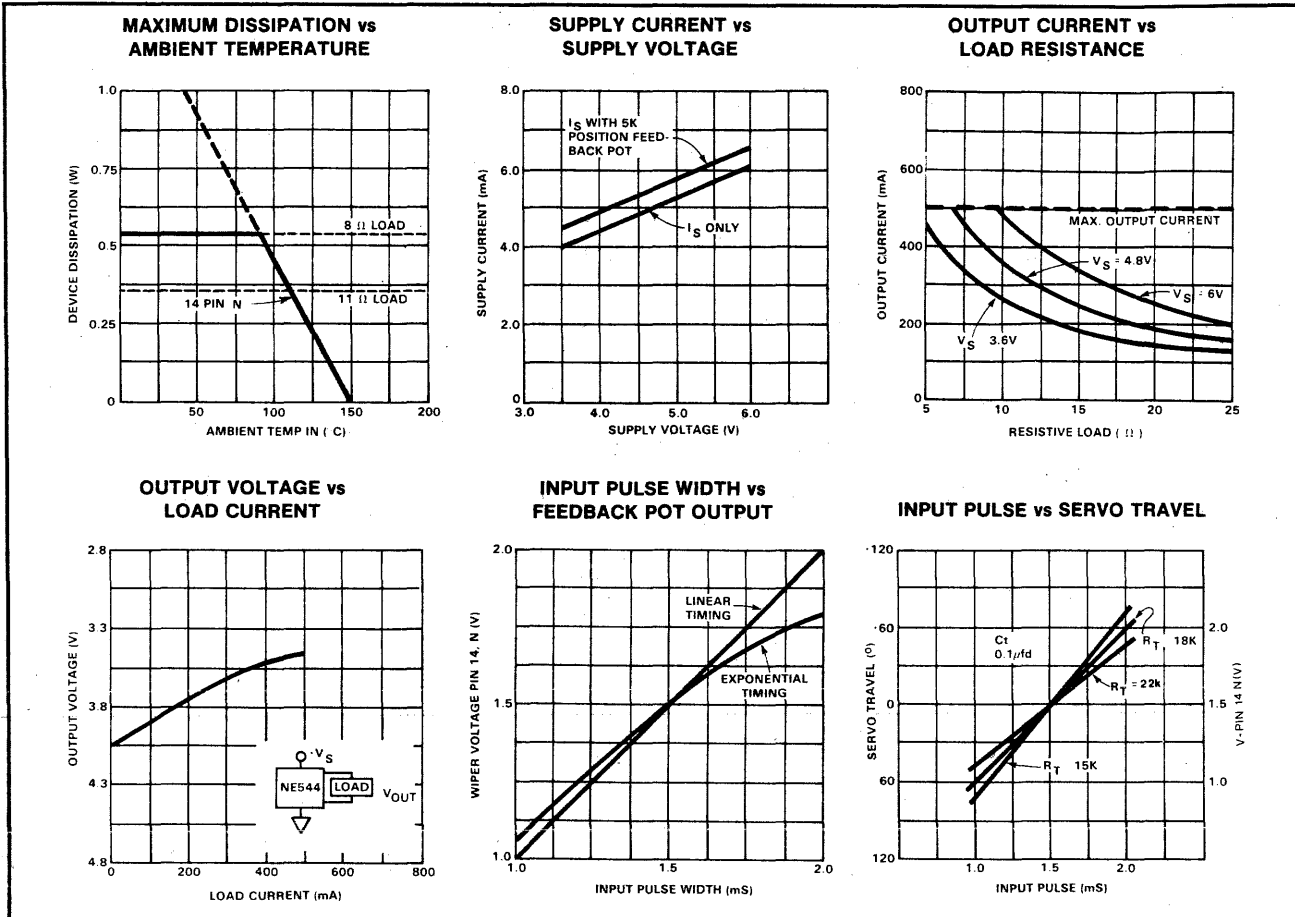
PARAMETER	TEST CONDITIONS	LIMITS			UNIT
		Min	Typ	Max	
$V_{CC}$ Supply voltage		3.2	4.8	6	V
$I_{CC}$ Supply current	Pin 11 Quiescent	4.2	5.5	10	mA
$V_{TH}$ Input threshold	Pin 4		1.5		V
			1.4		
$Z_{IN}$ Input resistance	Pin 4		18		k $\Omega$
$V_{OL}$ Output voltage	Pin 9 or 13, $I_L = 400\text{mA}$		0.3		V
$V_{OH}$ High			3.9		
$V_{REG}$ Regulated voltage	Pin 3	2.1	2.5	2.9	V
$\Delta V_{REG}$ Regulation	Pin 3 $3.9\text{V} \leq V_{CC} \leq 6\text{V}$ $R_{DB} = 0$		10		mV/V
	Pin 7 Minimum dead band		1		$\mu\text{s}$
	One shot temperature coefficient		.01		%/ $^\circ\text{C}$
	Pin 9 and 13 Standby output voltage		2.5		V
	Pin 10 and 12 PNP drive current		20		mA

LINEAR  
Signetics

SERVO AMPLIFIER

NE544

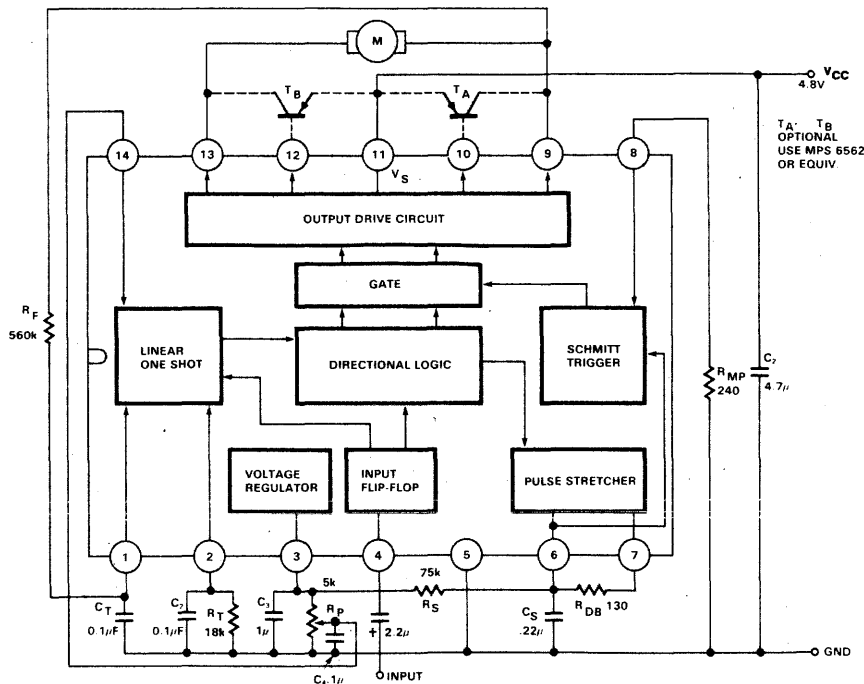
TYPICAL PERFORMANCE CHARACTERISTICS



LINEAR

Signetics

TYPICAL CONNECTION OF NE544N FOR LINEAR ONE SHOT TIMING



**LVDT SIGNAL CONDITIONER**

**NE5520**

**DESCRIPTION**

The NE5520 is a signal conditioning circuit for use with Linear Variable Differential Transformers (LVDT). The chip includes a low distortion amplitude stable sine wave oscillator with programmable frequency to drive the primary of the LVDT; a synchronous demodulator to convert the LVDT output amplitude and phase to position information; and an output amp to provide gain and filtering.

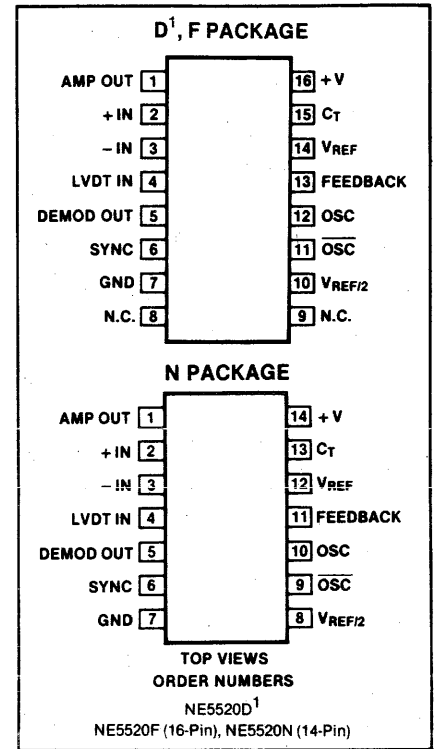
**FEATURES**

- Oscillator frequency: 1kHz to 20kHz
- Low distortion
- Capable of ratiometric operation
- Single supply operation 5V to 20V or dual supply  $\pm 2.5V$  to  $\pm 10V$
- Low power consumption

**APPLICATIONS**

- LVDT signal conditioning
- RVDT signal conditioning

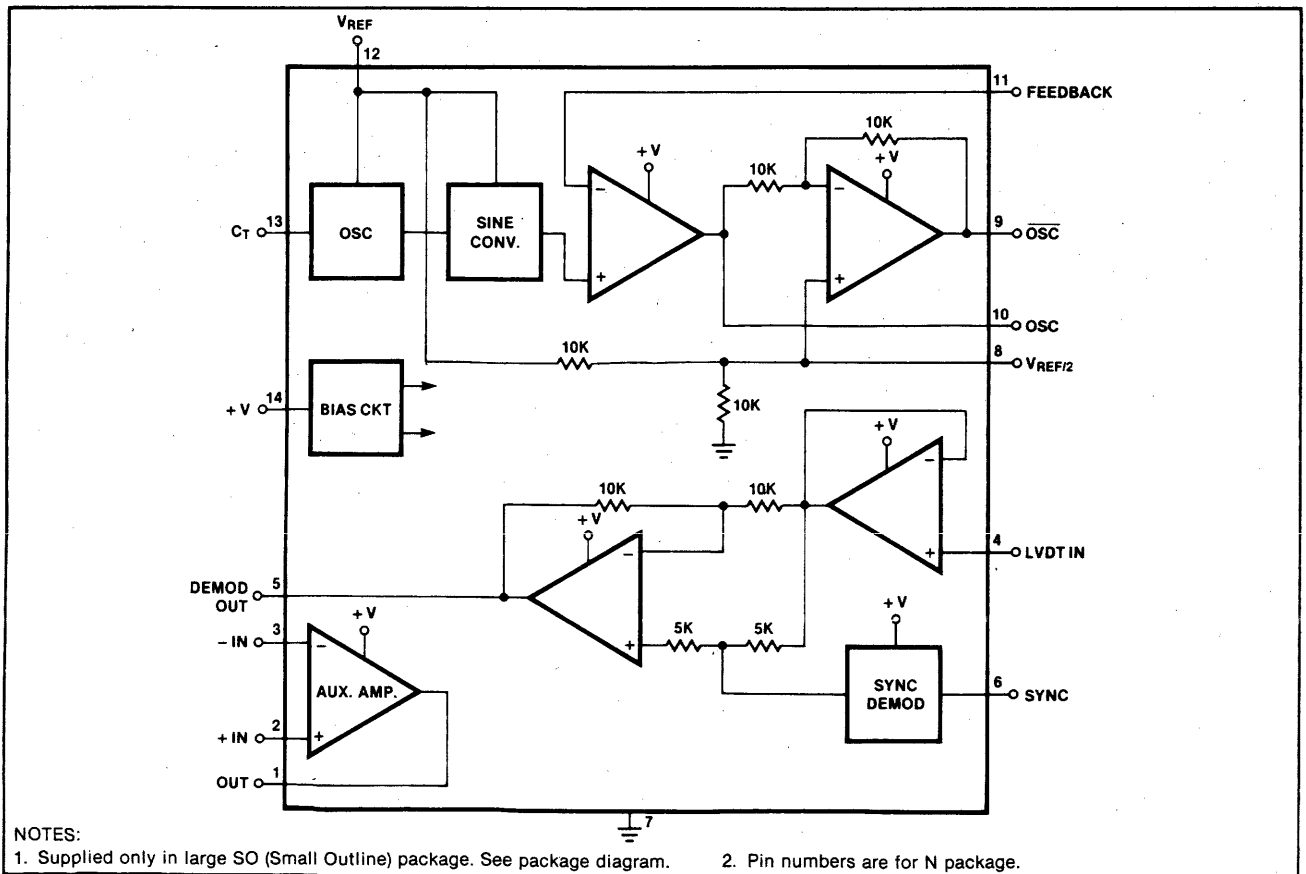
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
Supply voltage	+ 20	V
Split supply voltage	$\pm 10$	V
Operating temperature range	0 to + 70	$^{\circ}C$
Storage temperature range	- 65 to + 165	$^{\circ}C$
Power Dissipation (Note 1)	840	mW

**BLOCK DIAGRAM**



**OPERATIONAL TRANSCONDUCTANCE AMPLIFIERS**

**LM13600/13600A**

**DESCRIPTION**

The LM13600 series consists of two current controlled transconductance amplifiers each with differential inputs and a push pull output. The two amplifiers share common supplies but otherwise operate independently. Linearizing diodes are provided at the inputs to reduce distortion and allow higher input levels. The result is a 10 dB signal-to-noise improvement referenced to 0.5 percent THD. Controlled impedance buffers are provided which are specifically designed to complement the dynamic range of the amplifiers.

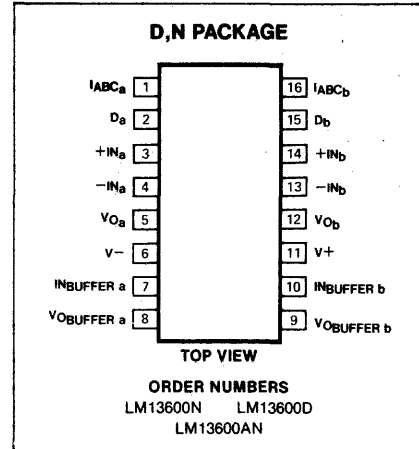
**FEATURES**

- gm adjustable over 6 decades
- Excellent gm linearity
- Excellent matching between amplifiers
- Linearizing diodes
- Controlled impedance buffers
- High output signal to noise ratio
- Wide supply range  $\pm 2V$  to  $\pm 22V$ .

**APPLICATIONS**

- Current controlled amplifiers
- Current controlled impedances
- Current controlled filters
- Current controlled oscillators
- Multiplexers
- Timers
- Sample and hold circuits

**PIN DESCRIPTION**

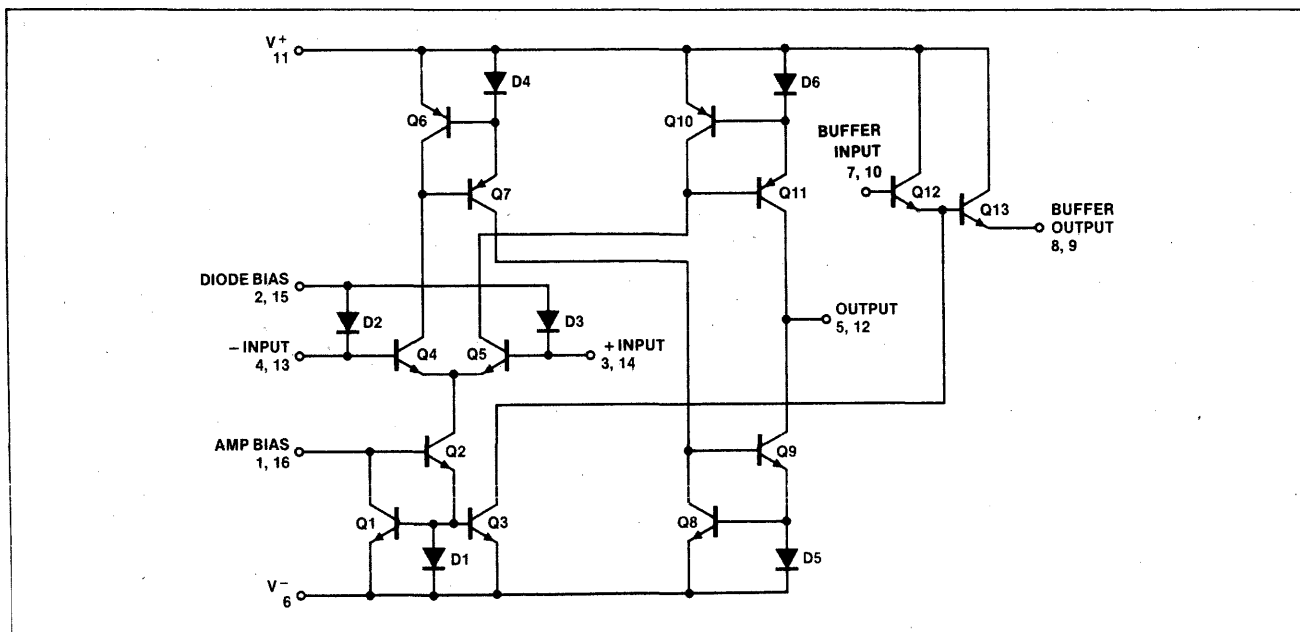


NOTE  
See Signetics NE5517 for typical circuit applications information.

**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
Supply voltage <sup>1</sup>		
LM13600	36 V <sub>DC</sub> or $\pm 18$	V
LM13600A	44 V <sub>DC</sub> or $\pm 22$	V
Power dissipation <sup>2</sup> T <sub>A</sub> = 25°C		
LM13600N, LM13600AN	570	mW
Differential input voltage	$\pm 5$	V
Diode bias current (I <sub>D</sub> )	2	mA
Amplifier bias current (I <sub>ABC</sub> )	2	mA
Output short circuit duration	Indefinite	
Buffer output current <sup>3</sup>	20	mA
Operating temperature range		
LM13600N, LM13600AN	0°C to +70	°C
DC input voltage	+V <sub>S</sub> to -V <sub>S</sub>	
Storage temperature range	-65°C to +150	°C
Lead temperature (Soldering, 10 Seconds)	300	°C

**BLOCK DIAGRAM**



LINEAR

Signetics

## OPERATIONAL TRANSCONDUCTANCE AMPLIFIERS

## LM13600/13600A

## ELECTRICAL CHARACTERISTICS

PARAMETER	TEST CONDITIONS	LM13600			LM13600A			UNIT
		Min	Typ	Max	Min	Typ	Max	
Input offset voltage ( $V_{OS}$ )	Over specified temperature range $I_{ABC} = 5\mu A$		0.4	5		0.4	2	mV
			0.3	5		0.3	5	mV
							2	mV
$\Delta V_{OS}/\Delta T$	Over temperature		7		7		$\mu V/^\circ C$	
$V_{OS}$ including diodes	Diode Bias Current ( $I_D$ ) = 500 $\mu A$		0.5	5		0.5	2	mV
Input offset change	$5\mu A \leq I_{ABC} \leq 500\mu A$		0.1			0.1	3	mV
Input offset current			0.1	0.6		0.1	0.6	$\mu A$
$\Delta I_{OS}/\Delta T$	Over temperature		0.001			0.001		$\mu A/^\circ C$
Input bias current	Over temperature range		0.4	5		0.4	5	$\mu A$
			1	8		1	7	$\mu A$
$\Delta I_B/\Delta T$	Over temperature		0.01			0.01		$\mu A/^\circ C$
Forward Transconductance (gm)	Over temperature range	6700	9600	13000	7700	9600	12000	$\mu mho$
		5400			4000			$\mu mho$
gm tracking			0.3			0.3		dB
Peak output current	$R_L = 0, I_{ABC} = 5\mu A$ $R_L = 0, I_{ABC} = 500\mu A$ $R_L = 0, \text{Over temp range}$		5		3	5	7	$\mu A$
		350	500	650	350	500	650	$\mu A$
		300			300			$\mu A$
Peak output voltage positive negative	$R_L = \infty, 5\mu A \leq I_{ABC} \leq 500\mu A$ $R_L = \infty, 5\mu A \leq I_{ABC} \leq 500\mu A$	+12	+14.2		+12	+14.2		V
		-12	-14.4		-12	-14.4		V
Supply current	$I_{ABC} = 500\mu A, \text{Both channels}$		2.6	4		2.6	4	mA
$V_{OS}$ sensitivity positive negative	$\Delta V_{OS}/\Delta V+$ $\Delta V_{OS}/\Delta V-$		20	150		20	150	$\mu V/V$
			20	150		20	150	$\mu V/V$
CMRR		80	110		80	110		dB
Common mode range		$\pm 12$	$\pm 13.5$		$\pm 12$	$\pm 13.5$		V
Crosstalk	Referred to input <sup>5</sup> 20 Hz < f < 20 KHz		100			100		dB
Diff. input current	$I_{ABC} = 0, \text{Input} = \pm 4V$		0.02	100		0.02	10	nA
Leakage current	$I_{ABC} = 0$ (Refer to test circuit)		0.2	100		0.2	5	nA
Input resistance		10	26		10	26		K $\Omega$
Open loop bandwidth			2			2		MHz
Slew rate	Unity gain compensated		50			50		V/ $\mu$ Sec
Buff. input current	5		0.4	5		0.4	5	$\mu A$
Peak buffer output voltage	5	10			10			V

## NOTES

- For selections to a supply voltage above  $\pm 22V$ , contact factory.
- For operating at high temperatures, the device must be derated based on a  $150^\circ C$  maximum junction temperature and a thermal resistance of  $175^\circ C/W$  which applies for the device soldered in a printed circuit board, operating in still air.
- Buffer output current should be limited so as to not exceed package dissipation.
- These specifications apply for  $V_S = \pm 15V, T_A = 25^\circ C$ , amplifier bias current ( $I_{ABC}$ ) =  $500\mu A$ , pins 2 and 15 open unless otherwise specified. The inputs to the buffers are grounded and outputs are open.
- These specifications apply for  $V_{SS} = \pm 15V, I_{ABC} = 500\mu A, R_{OUT} = 5K\Omega$  connected from the buffer output to  $-V_S$  and the input of the buffer is connected to the transconductance amplifier output.

DUAL HIGH PERFORMANCE OPERATIONAL AMPLIFIER

NE/SE5512

DESCRIPTION

The 5512 series of high performance operational amplifier provides very good input characteristics. These amplifiers feature low input bias and voltage characteristics such as a 108 op amp with improved CMRR and a high differential input voltage limit achieved through the use of a bias cancellation and PNP input circuits with collector to emitter clamping. The output characteristics are like those of a 741 op amp with improved slew rate and drive capability yet have low supply quiescent current.

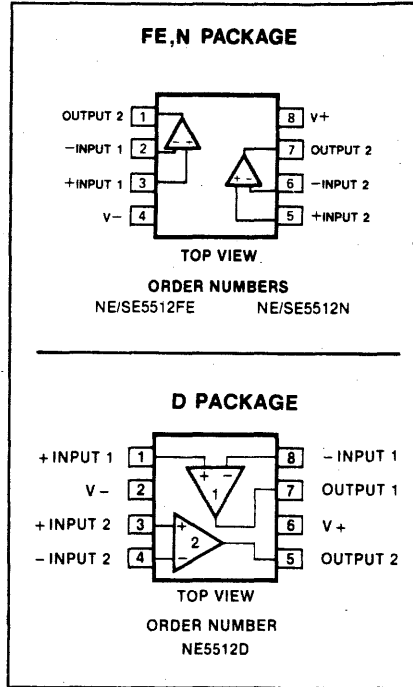
APPLICATIONS

- AC amplifiers
- RC active filters
- Transducer amplifiers
- DC gain block
- Battery operation
- Instrumentation amplifiers

FEATURES

- Low input bias  $< \pm 3nA$
- Low input offset current  $< \pm 3nA$
- Low input offset voltage  $< 1mV$
- Low  $V_{OS}$  temperature drift  $4\mu V/^{\circ}C$
- Low input bias temperature drift  $30pA/^{\circ}C$
- Low input voltage noise  $25nV/\sqrt{Hz}$
- Low supply current  $1.5mA/amp$
- High slew rate  $1.0V/\mu s$
- High CMRR 100dB
- High input impedance  $100M\Omega$
- High PSRR 110dB
- High differential input voltage limit
- No cross-over distortion
- Indefinite output short circuit protection
- Internally compensated for unity gain

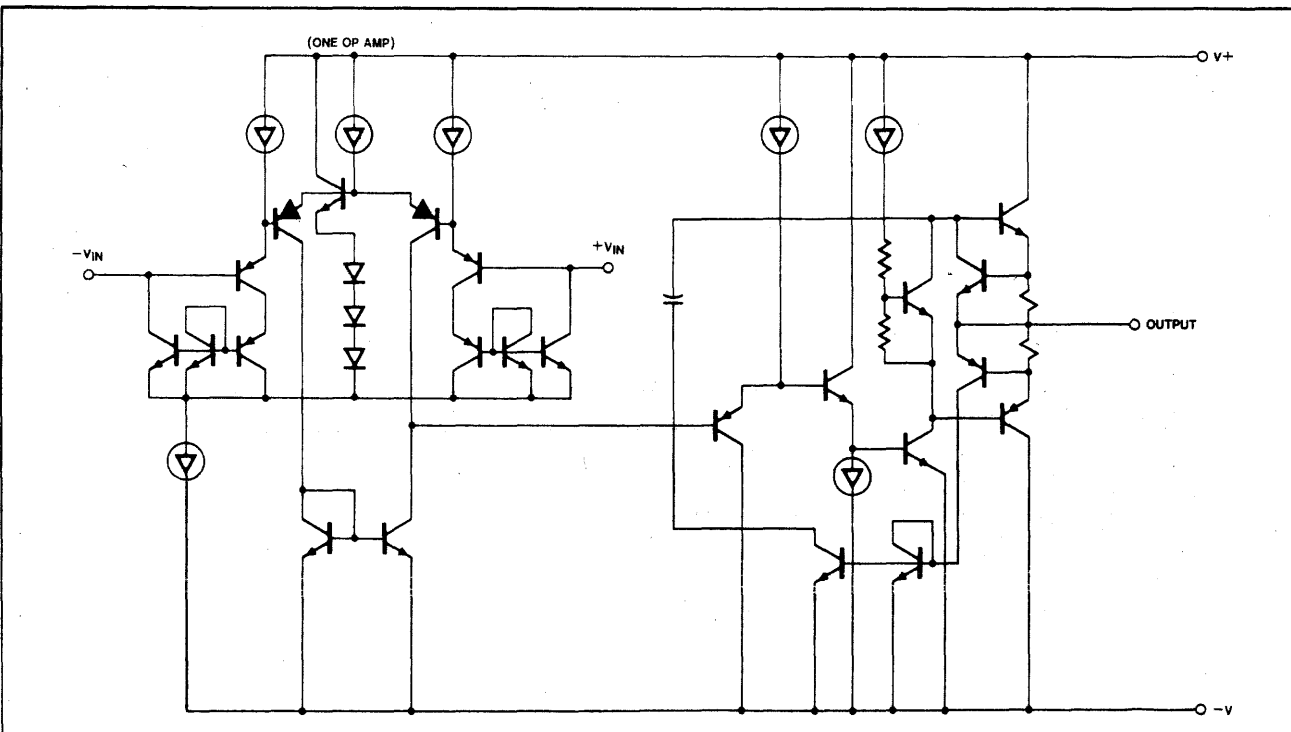
PIN CONFIGURATIONS



ABSOLUTE MAXIMUM RATINGS

Parameter	Rating	Unit
V <sub>CC</sub> Supply Voltage	$\pm 16$	V
V <sub>D</sub> Power dissipation	500	mW
T <sub>A</sub> Operating temperature range		
NE5512	0 to 70	$^{\circ}C$
SE5512	-55 to +125	$^{\circ}C$
T <sub>STG</sub> Storage temperature range	-65 to +150	$^{\circ}C$
T <sub>SOLD</sub> Lead temperature soldering	300	$^{\circ}C$

EQUIVALENT SCHEMATIC



LINEAR

Signetics



# DUAL HIGH PERFORMANCE OPERATIONAL AMPLIFIER

## NE/SE5512

### ELECTRICAL PERFORMANCE CHARACTERISTICS $V_{CC} = \pm 15V$ , F.R. = $-55^{\circ}C$ to $+125^{\circ}C$ (SE), $0^{\circ}C$ to $+70^{\circ}C$ (NE)

	PARAMETER	TEST CONDITIONS	SE5512			NE5512			UNIT
			Min	Typ	Max	Min	Typ	Max	
$V_{OS}$	Input offset voltage	$R_S = 100\Omega$ $T_A = +25^{\circ}C$ $T_A = F.R.$		0.7	2		1	5	mV
$\Delta V_{OS}$		Over Temp.		1	3		1.5	6	$\mu V/^{\circ}C$
				4			5		
$I_{OS}$	Input offset current	$R_S = 100k\Omega$ $T_A = +25^{\circ}C$ $T_A = F.R.$		3	10		6	20	nA
$\Delta I_{OS}$		Over Temp.		4	20		8	30	$pA/^{\circ}C$
				30			40		
$I_B$	Input bias current	$R_S = 100k\Omega$ $T = +25^{\circ}C$ $T_A = F.R.$		3	10		6	20	nA
$\Delta I_B$		Over Temp.		4	20		8	30	$pA/^{\circ}C$
				30			40		
$R_{IN}$	Input resistance differential	$T_A = 25^{\circ}C$		100			100		M $\Omega$
$V_{CM}$	Input common mode range	$T_A = 25^{\circ}C$ $T_A = F.R.$	$\pm 13.5$ $\pm 13$	$\pm 13.7$ $\pm 13.2$		$\pm 13.5$ $\pm 13$	$\pm 13.7$ $\pm 13.2$		V
CMRR	Input common-mode rejection ratio	$V_{CC} = \pm 15V$ $V_{IN} = \pm 13.5V$ (RM) $T_A = 25^{\circ}C$ $V_{IN} = \pm 13V$ (F.R.) $T_A = F.R.$	70	100		70	100		dB
$A_{VOL}$	Large-signal voltage gain	$R_L = 2k\Omega$ $T_A = 25^{\circ}C$ $V_O = \pm 10V$ $T_A = F.R.$	50 25	200		50 25	200		V/mV
S.R.	Slew rate	$T_A = 25^{\circ}C$	0.6	1			1		V/ $\mu s$
GBW	Small-signal unity gain bandwidth	$T_A = 25^{\circ}C$		3			3		MHz
$\theta_M$	Phase margin	$T_A = 25^{\circ}C$		45			45		Degree
$V_{OUT}$	Output voltage swing	$R_L = 2k\Omega$ $T_A = 25^{\circ}C$ $T_A = F.R.$	$\pm 13$ $\pm 12.5$	$\pm 13.5$ $\pm 13$		$\pm 13$ $\pm 12.5$	$\pm 13.5$ $\pm 13$		V
$V_{OUT}$	Output voltage swing	$R_L = 600\Omega^*$ $T_A = 25^{\circ}C$ $T_A = F.R.$	$\pm 10$ $\pm 7.5$	$\pm 11.5$ $\pm 9$		$\pm 10$ $\pm 8$	$\pm 11.5$ $\pm 9$		V
$I_{CC}$	Power supply current	$R_L = \text{Open}$ $T_A = 25^{\circ}C$ $T_A = F.R.$		3.4 3.6	5 5.5		3.4 3.6	5 5.5	mA
$P_{SRR}$	Power supply rejection ratio	$T_A = 25^{\circ}C$ $T_A = F.R.$	80 80	110 100		80 80	110 100		dB
AA	Amplifier to amplifier coupling	$f = 1kHz$ to $20kHz$ $T_A = 25^{\circ}C$		-120			-120		dB
HD	Total harmonic distortion	$f = 10kHz$ $T_A = 25^{\circ}C$ $V_O = 7V_{RMS}$		0.01			0.01		%
$V_{INN}$	Input noise voltage	$f = 1kHz$ $T_A = 25^{\circ}C$		30			30		nV/ $\sqrt{Hz}$
$I_{INN}$	Input noise current	$f = 1kHz$ $T_A = 25^{\circ}C$		.2			.2		$pA/\sqrt{Hz}$
$I_{SC}$	Short circuit	$\pm 15V$ $T_A = 25^{\circ}C$		40			40		mA

NOTE

For operation at elevated temperature, N package must be derated based on a thermal resistance of  $120^{\circ}W$  junction to ambient. Thermal resistance of the FE package is  $125^{\circ}W$ .

**QUAD HIGH PERFORMANCE OP AMP**

**NE/SE5514**

**DESCRIPTION**

The NE/SE5514 family of Quad Operational Amplifiers sets new standards in Bipolar Quad Amplifier Performance. The amplifiers feature low input bias current and low offset voltages. Pin-out is identical to LM324/LM348 which facilitates direct product substitution for improved system performance. Output characteristics are similar to a  $\mu$ A741 with improved slew and drive capability.

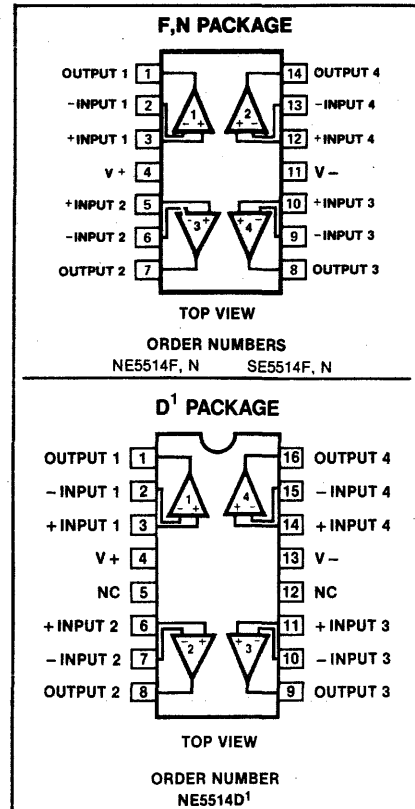
**FEATURES**

- Low input bias current:  $< \pm 3nA$
- Low input offset current:  $< \pm 3nA$
- Low input offset voltage:  $< 1mV$
- Low supply current: 1.5mA/Amp
- 1 V/ $\mu$ sec slew rate
- High input impedance: 100M $\Omega$
- High common mode impedance: 10G $\Omega$
- Internal compensation for unity gain

**APPLICATIONS**

- AC amplifiers
- RC active filters
- Transducer amplifiers
- DC gain block
- Instrumentation amplifier

**PIN CONFIGURATION**

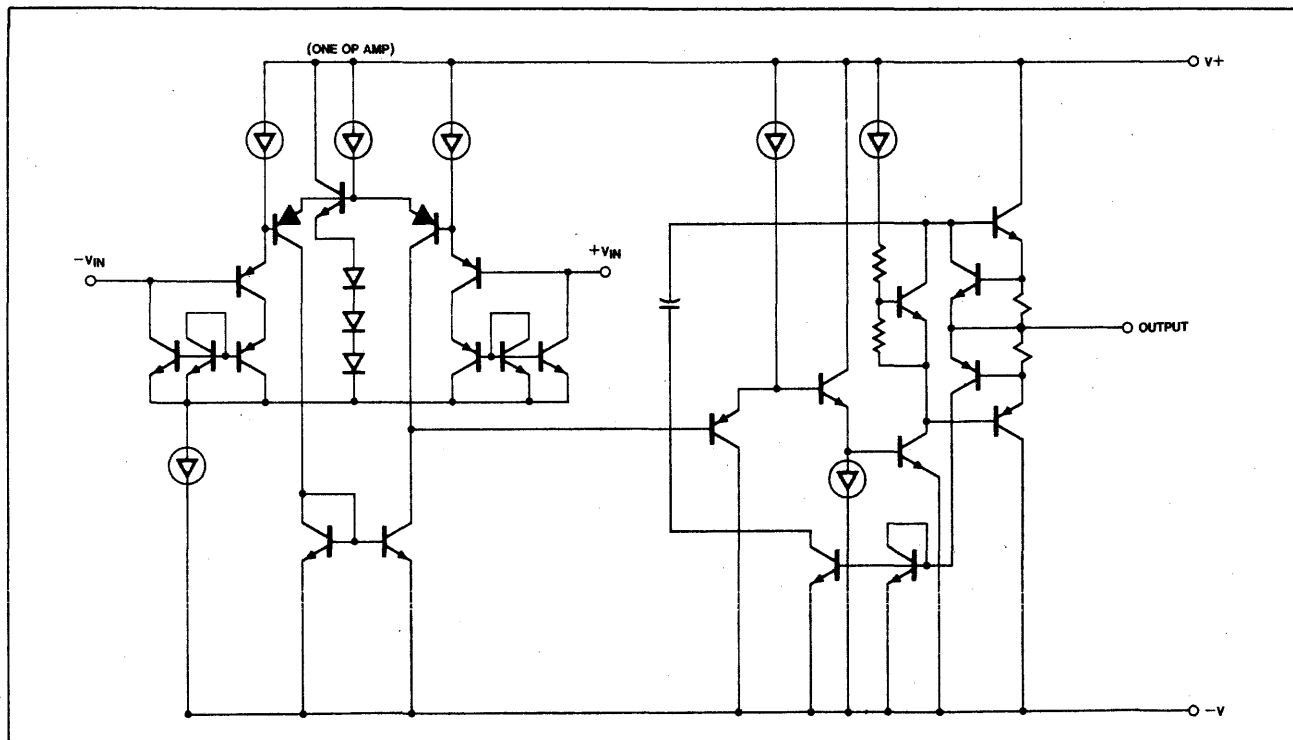


**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub>	Supply voltage	$\pm 16$ V
V <sub>DIFF</sub>	Differential input voltage	32 V
V <sub>IN</sub>	Input voltage	0 to 32 V
	Output short to ground	Continuous
T <sub>S</sub>	Storage temperature range	-65 to +150 °C
T <sub>SOLD</sub>	Lead soldering temperature	300 °C
T <sub>A</sub>	Operating temperature range	
	NE5514	0 to 70 °C
	SE5514	-55 to +125 °C

NOTE:  
1. Supplied only in large SO (Small Outline) package.

**EQUIVALENT SCHEMATIC**



LINEAR

Signetics

# QUAD HIGH PERFORMANCE OP AMP

NE/SE5514

## ELECTRICAL CHARACTERISTICS $V_{CC} = \pm 15V$ , F.R. = $-55^{\circ}C$ (SE) $0^{\circ}C$ to $70^{\circ}C$ (NE)

PARAMETER	TEST CONDITIONS	SE5514			NE5514			UNIT		
		Min	Typ	Max	Min	Typ	Max			
$V_{OS}$	Input offset voltage		0.7	2		1	5	mV		
$\Delta V_{OS}$			1	3		1.5	6	$\mu V/^{\circ}C$		
$I_{OS}$	Input offset current		3	10		6	20	nA		
$\Delta I_{OS}$			4	20		8	30	pA/ $^{\circ}C$		
$I_B$	Input bias current		3	10		6	20	nA		
$\Delta I_B$			4	20		8	30	pA/ $^{\circ}C$		
$R_{IN}$	Input resistance differential		100			100		M $\Omega$		
$V_{CM}$	Input common mode range		$\pm 13.5$	$\pm 13.7$		$\pm 13.5$	$\pm 13.7$	V		
			$\pm 13.5$	$\pm 13.2$		$\pm 13$	$\pm 13.2$			
CMRR	Input common-mode rejection ratio		70	100		70	100	dB		
AVOL	Large-signal voltage gain		50			50		V/mV		
GAIN			25	200		25				
S.R.	Slew rate		0.6	1		0.6	1	V/ $\mu s$		
GBW	Small-signal unity gain bandwidth			3			3	MHz		
$\theta_M$	Phase margin			45			45	Degr		
$V_{OUT}$	Output voltage swing		$\pm 13$	$\pm 13.5$		$\pm 13$	$\pm 13.5$	V		
			$\pm 12.5$	$\pm 13$		$\pm 12.5$	$\pm 13$			
$V_{OUT}$	Output voltage swing		$\pm 10$	$\pm 11.5$		$\pm 10$	$\pm 11.5$	V		
			$\pm 7.5$	$\pm 9$		$\pm 8$	$\pm 9$			
$I_{CC}$	Power supply current			6		6	10	mA		
				7		7	12			
PSRR	Power supply rejection ratio		80	110		80	110	dB		
			80	100		80	100			
AA	Amplifier to amplifier coupling			-120			-120	dB		
HD	Total harmonic distortion			0.01			0.01	%		
$V_{INN}$	Input-noise voltage			30			30	$nV/\sqrt{Hz}$		
$I_{SC}$	Short Circuit		10	40		60	10	40	60	mA

NOTE

\*For operation at elevated temperature, N package must be derated based on a thermal resistance of 95 $^{\circ}C/W$  junction to ambient.

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DUAL OPERATIONAL TRANSCONDUCTANCE AMPLIFIER

NE5517/5517A

DESCRIPTION

The NE5517 contains two current controlled transconductance amplifiers, each with a differential input and push-pull output. The NE5517 offers significant design and performance advantages over similar devices for all types of programmable gain applications. Constant impedance buffers are provided which effectively eliminate changes in output offset voltage as the amplifier bias current is varied. Circuit performance is enhanced through the use of linearizing diodes at the inputs which enable a 10 dB signal to noise improvement referenced to .5 percent THD. The NE5517 is suited for a wide variety of industrial and consumer applications and is recommended as the preferred circuit in the Dolby® HX (Headroom Extension) system.

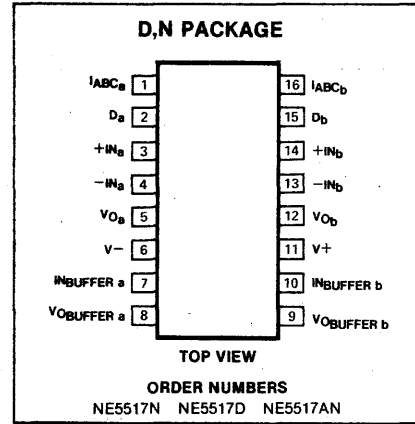
FEATURES

- Constant impedance buffers
- $\Delta V_{BE}$  of buffer is constant with amplifier  $I_{BIAS}$  change
- Pin compatible with LM13600
- Excellent matching between amplifiers
- Linearizing diodes
- High output signal-to-noise ratio

APPLICATIONS

- Multiplexers
- Timers
- Electronic music synthesizers
- Dolby® HX Systems
- Current-controlled amplifiers, filters
- Current-controlled oscillators, impedances

PIN CONFIGURATION

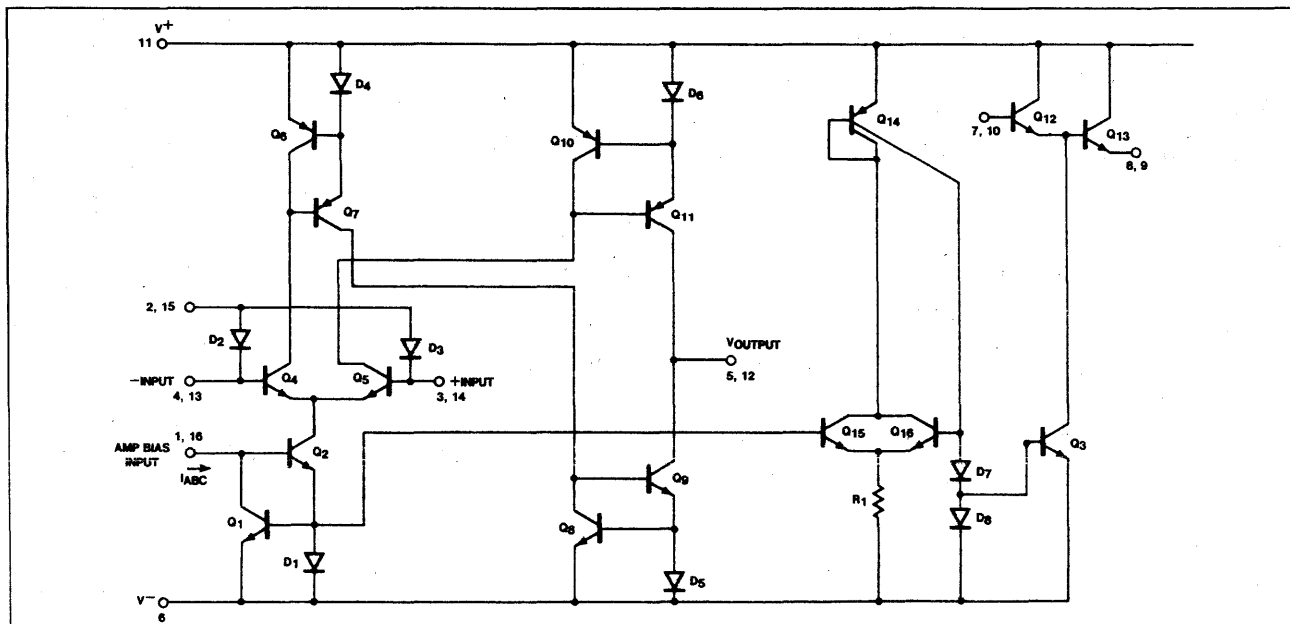


NOTE  
 \*Dolby is a registered trademark of Dolby Laboratories Inc., San Francisco, Calif.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Supply Voltage <sup>1</sup>		
NE5517	36 V <sub>DC</sub> or $\pm 18$	V
NE5517A	44 V <sub>DC</sub> or $\pm 22$	V
Power Dissipation <sup>2</sup> T <sub>A</sub> = 25°C		
NE5517N, NE5517AN	570	mW
Differential Input Voltage	$\pm 5$	V
Diode Bias Current (I <sub>D</sub> )	2	mA
Amplifier Bias Current (I <sub>ABC</sub> )	2	mA
Output Short Circuit Duration	Indefinite	
Buffer Output Current <sup>3</sup>	20	mA
Operating Temperature Range		
NE5517N, NE5517AN	0°C to +70	°C
DC Input Voltage	+V <sub>S</sub> to -V <sub>S</sub>	
Storage Temperature Range	-65°C to +150	°C
Lead Temperature (Soldering, 10 Seconds)	300	°C

CIRCUIT SCHEMATIC



LINEAR  
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# DUAL OPERATIONAL TRANSCONDUCTANCE AMPLIFIER

## NE5517/5517A

### ELECTRICAL CHARACTERISTICS<sup>4</sup>

PARAMETER	TEST CONDITIONS	NE5517			NE5517A			UNIT
		Min	Typ	Max	Min	Typ	Max	
Input offset voltage ( $V_{OS}$ )	Over temperature range $I_{ABC} = 5\mu A$		0.4	5		0.4	2	mV
			0.3	5		0.3	5	mV
$\Delta V_{OS}/\Delta T$	Avg. TC of input offset voltage		7			7		$\mu V/^\circ C$
$V_{OS}$ including diodes	Diode bias current ( $I_D$ ) = 500 $\mu A$		0.5	5		0.5	2	mV
Input offset change	$5\mu A \leq I_{ABC} \leq 500\mu A$		0.1			0.1	3	mV
Input offset current			0.1	0.6		0.1	0.6	$\mu A$
$\Delta I_{OS}/\Delta T$	Avg. TC of input offset current		0.001			0.001		$\mu A/^\circ C$
Input bias current	Over temperature range		0.4	5		0.4	5	$\mu A$
			1	8		1	7	$\mu A$
$\Delta I_B/\Delta T$	Avg. TC of input current		0.01			0.01		$\mu A/^\circ C$
Forward Transconductance (gm)	Over temperature range	6700	9600	13000	7700	9600	12000	$\mu mho$
		5400			4000			$\mu mho$
gm tracking			0.3			0.3		dB
Peak output current	$R_L = 0, I_{ABC} = 5\mu A$ $R_L = 0, I_{ABC} = 500\mu A$ $R_L = 0,$		5		3	5	7	$\mu A$
		350	500	650	350	500	650	$\mu A$
		300			300			$\mu A$
Peak output voltage	$R_L = \infty, 5\mu A \leq I_{ABC} \leq 500\mu A$ $R_L = \infty, 5\mu A \leq I_{ABC} \leq 500\mu A$	+12	+14.2		+12	+14.2		V
		-12	-14.4		-12	-14.4		V
Supply current	$I_{ABC} = 500\mu A$ , both channels		2.6	4		2.6	4	mA
$V_{OS}$ sensitivity	$\Delta V_{OS}/\Delta V+$ $\Delta V_{OS}/\Delta V-$		20	150		20	150	$\mu V/V$
			20	150		20	150	$\mu V/V$
CMRR		80	110		80	110		dB
Common mode range		$\pm 12$	$\pm 13.5$		$\pm 12$	$\pm 13.5$		V
Crosstalk	Referred to input <sup>5</sup> 20Hz < f < 20kHz		100			100		dB
Diff. input current	$I_{ABC} = 0$ , input = $\pm 4V$		0.02	100		0.02	10	nA
Leakage current	$I_{ABC} = 0$ (Refer to test circuit)		0.2	100		0.2	5	nA
Input resistance		10	26		10	26		K $\Omega$
Open loop bandwidth			2			2		MHz
Slew rate	Unity gain compensated		50			50		V/ $\mu Sec$
Buff. input current	5		0.4	5		0.4	5	$\mu A$
Peak buffer output voltage	5	10			10			V
$\Delta V_{BE}$ of buffer	6 Refer to Buffer $V_{BE}$ test circuit		0.5	5		0.5	5	mV

NOTES

- For selections to a supply voltage above  $\pm 22V$ , contact factory.
- For operating at high temperatures, the device must be derated based on a 150 $^\circ C$  maximum junction temperature and a thermal resistance of 175 $^\circ C/W$  which applies for the device soldered in a printed circuit board, operating in still air.
- Buffer output current should be limited so as to not exceed package dissipation.
- These specifications apply for  $V_S = \pm 15V$ ,  $T_A = 25^\circ C$ , amplifier bias current ( $I_{ABC}$ ) = 500 $\mu A$ , pins 2 and 15 open unless otherwise specified. The inputs to the buffers are

grounded and outputs are open.

- These specifications apply for  $V_S = \pm 15V$ ,  $I_{ABC} = 500\mu A$ ,  $R_{OUT} = 5k\Omega$  connected from the buffer output to  $-V_S$  and the input of the buffer is connected to the transconductance amplifier output.
- $V_S = \pm 15$ ,  $R_{OUT} = 5k\Omega$  connected from Buffer output to  $-V_S$  and  $5\mu A \leq I_{ABC} \leq 500\mu A$ .

LINEAR  
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**INTERNALLY COMPENSATED DUAL LOW NOISE OP AMP NE/SE5532/5532A**

**DESCRIPTION**

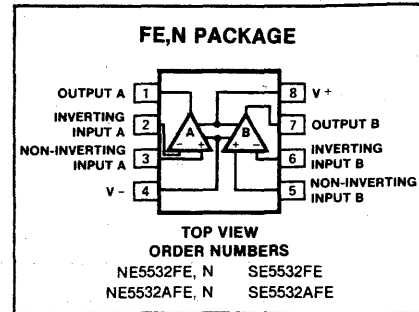
The 5532 is a dual high-performance low noise operational amplifier. Compared to most of the standard operational amplifiers, such as the 1458, it shows better noise performance, improved output drive capability and considerably higher small-signal and power bandwidths.

This makes the device especially suitable for application in high quality and professional audio equipment, instrumentation and control circuits, and telephone channel amplifiers. The op amp is internally compensated for gains equal to one. If very low noise is of prime importance, it is recommended that the 5532A version be used which has guaranteed noise voltage specifications.

**FEATURES**

- Small-signal bandwidth: 10MHz
- Output drive capability: 600Ω, 10V (rms)
- Input noise voltage: 5nV/√Hz (typical)
- DC voltage gain: 50000
- AC voltage gain: 2200 at 10kHz
- Power bandwidth: 140kHz
- Slew-rate: 9V/μs
- Large supply voltage range: ±3 to ±20V

**PIN CONFIGURATION**



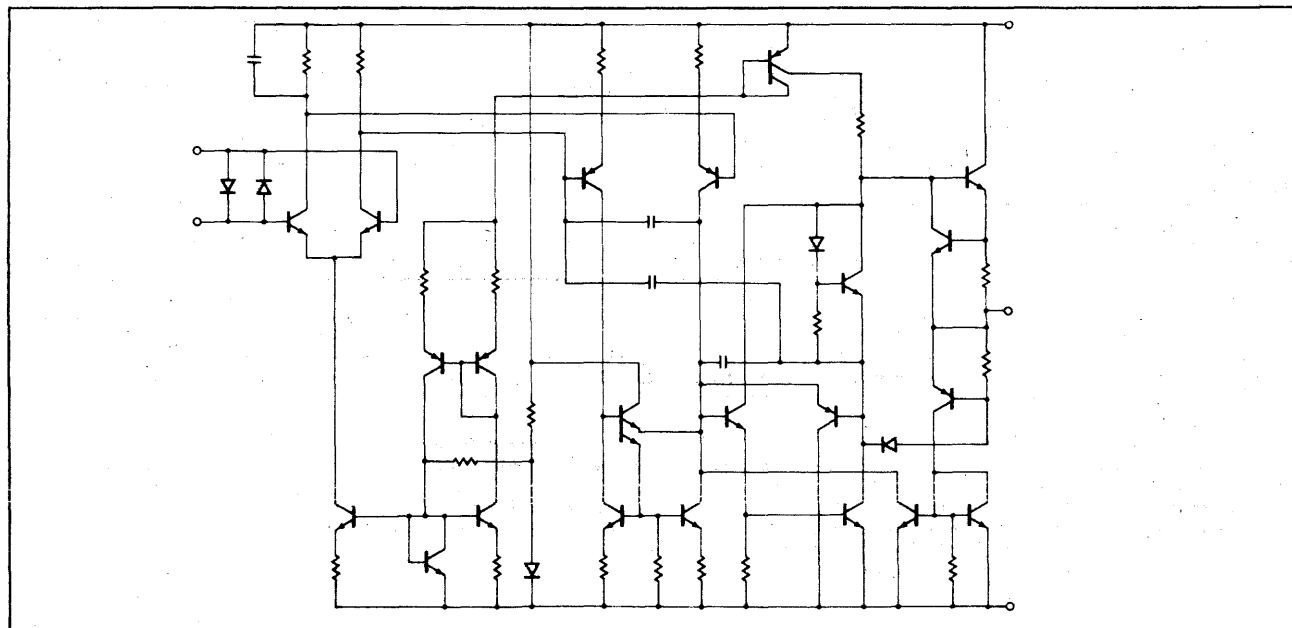
**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>S</sub> Supply voltage	±22	V
V <sub>IN</sub> Input voltage	±V supply	V
V <sub>DIFF</sub> Differential input voltage <sup>1</sup>	±.5	V
T <sub>A</sub> Operating temperature range	0 to 70	°C
T <sub>STG</sub> Storage temperature	-65 to +150	°C
T <sub>J</sub> Junction temperature	150	°C
P <sub>D</sub> Power dissipation		
5532FE	1000	mW
Lead temperature (soldering, 10 sec)	300	°C

**NOTES:**

1. Diodes protect the inputs against over-voltage. Therefore, unless current-limiting resistors are used, large currents will flow if the differential input voltage exceeds 0.6V. Maximum current should be limited to ±10mA.
2. Thermal resistance of the FE package is 125°C/W.

**EQUIVALENT SCHEMATIC (EACH AMPLIFIER)**



LINEAR

Signetics

# INTERNALLY COMPENSATED DUAL LOW NOISE OP AMP NE/SE5532/5532A

## DC ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$ , $V_S = \pm 15\text{V}$ unless otherwise specified.<sup>1,2</sup>

PARAMETER	TEST CONDITIONS	SE5532/55232A			NE5532/5532A			UNIT
		Min	Typ	Max	Min	Typ	Max	
$V_{OS}$ Offset voltage	Over temperature		0.5	2		0.5	4	mV
$\Delta V_{OS}/\Delta T$			5	3		5	5	mV/ $\mu\text{V}/^\circ\text{C}$
$I_{OS}$ Offset current	Over temperature			100		10	150	nA
$\Delta I_{OS}/\Delta T$			200	200		200	200	nA/ $\text{pA}/^\circ\text{C}$
$I_B$ Input current	Over temperature		200	400		200	800	nA
$\Delta I_B/\Delta T$			5	700		5	1000	nA/ $\text{mA}/^\circ\text{C}$
$I_{CC}$ Supply current	Over temperature			13		8	16	mA mA
$V_{CM}$ Common mode input range		$\pm 12$	$\pm 13$		$\pm 12$	$\pm 13$		V
CMRR Common mode rejection ratio		80	100		70	100		dB
PSRR Power supply rejection ratio			10	50		10	100	$\mu\text{V}/\text{V}$
$A_{VOL}$ Large signal voltage gain	$R_L \geq 2\text{k}\Omega$ , $V_0 = \pm 10\text{V}$ Over temperature $R_L \geq 600\Omega$ , $V_0 = \pm 10\text{V}$ Over temperature	50			25	100		V/mV
		25			15			V/mV
		40			15	50		V/mV
		20			10			V/mV
$V_{OUT}$ Output swing	$R_L \geq 600\Omega$ Over temperature $R_L \geq 600\Omega$ , $V_S = \pm 18\text{V}$ Over temperature $R_L \geq 2\text{k}\Omega$ over temp.				$\pm 12$	$\pm 13$		V
					$\pm 10$	$\pm 12$		V
		$\pm 15$	$\pm 16$		V			V
		$\pm 12$	$\pm 13$		$\pm 12$	$\pm 14$		V
$R_{IN}$ Input resistance		30	300		30	300		k $\Omega$
$I_{SC}$ Output short circuit current		10	38	60	10	38	60	mA

NOTES

1. For NE5532, NE5532A,  $T_{Min} = 0^\circ\text{C}$ ,  $T_{Max} = 70^\circ\text{C}$ .

2. For SE5532/55232A,  $T_{Min} = -55^\circ\text{C}$ ,  $T_{Max} = +125^\circ\text{C}$ .

## AC ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$ , $V_S = \pm 15\text{V}$ unless otherwise specified.

PARAMETER	TEST CONDITIONS	NE/SE5532/5532A			UNIT
		Min	Typ	Max	
$R_{OUT}$ Output resistance	$A_V = 30\text{dB}$ Closed loop $f = 10\text{kHz}$ , $R_L = 600\Omega$		0.3		$\Omega$
Overshoot	Voltage follower $V_{IN} = 100\text{mV}$ p-p $C_L = 100\text{pF}$ $R_L = 600\Omega$		10		%
Gain	$f = 10\text{kHz}$		2.2		V/mV
Gain bandwidth product	$C_L = 100\text{pF}$ $R_L = 600\Omega$		10		MHz
Slew rate			9		V/ $\mu\text{s}$
Power bandwidth	$V_{OUT} = \pm 10\text{V}$ $V_{OUT} = \pm 14\text{V}$ , $R_L = 600\Omega$ , $V_{CC} = \pm 18\text{V}$		140		kHz
			100		kHz

## ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$ , $V_S = \pm 15\text{V}$ unless otherwise specified.

PARAMETER	TEST CONDITIONS	NE/SE5532			NE/SE5532A			UNIT
		Min	Typ	Max	Min	Typ	Max	
Input noise voltage	$f_o = 30\text{Hz}$		8			8	12	$\text{nV}/\sqrt{\text{Hz}}$
	$f_o = 1\text{kHz}$		5			5	6	$\text{nV}/\sqrt{\text{Hz}}$
Input noise current	$f_o = 30\text{Hz}$		2.7			2.7		$\text{pA}/\sqrt{\text{Hz}}$
	$f_o = 1\text{kHz}$		0.7			0.7		$\text{pA}/\sqrt{\text{Hz}}$
Channel separation	$f = 1\text{kHz}$ , $R_S = 5\text{k}\Omega$		110			110		dB

LINEAR  
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**SINGLE AND DUAL LOW NOISE OP AMP NE5533/5533A / NE/SE5534/5534A**

**DESCRIPTION**

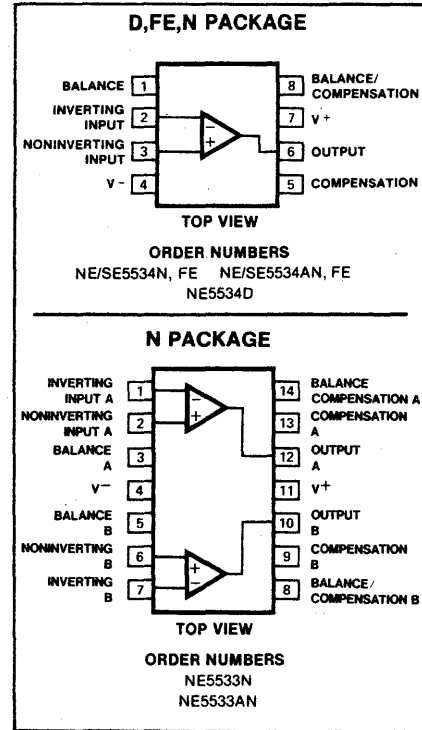
The 5533/5534 are dual and single high-performance low noise operational amplifiers. Compared to other operational amplifiers, such as TL083, they show better noise performance, improved output drive capability and considerably higher small-signal and power bandwidths.

This makes the devices especially suitable for application in high quality and professional audio equipment, in instrumentation and control circuits and telephone channel amplifiers. The op amps are internally compensated for gain equal to, or higher than, three. The frequency response can be optimized with an external compensation capacitor for various applications (unity gain amplifier, capacitive load, slew-rate, low overshoot, etc.) If very low noise is of prime importance, it is recommended that the 5533A/5534A version be used which has guaranteed noise specifications.

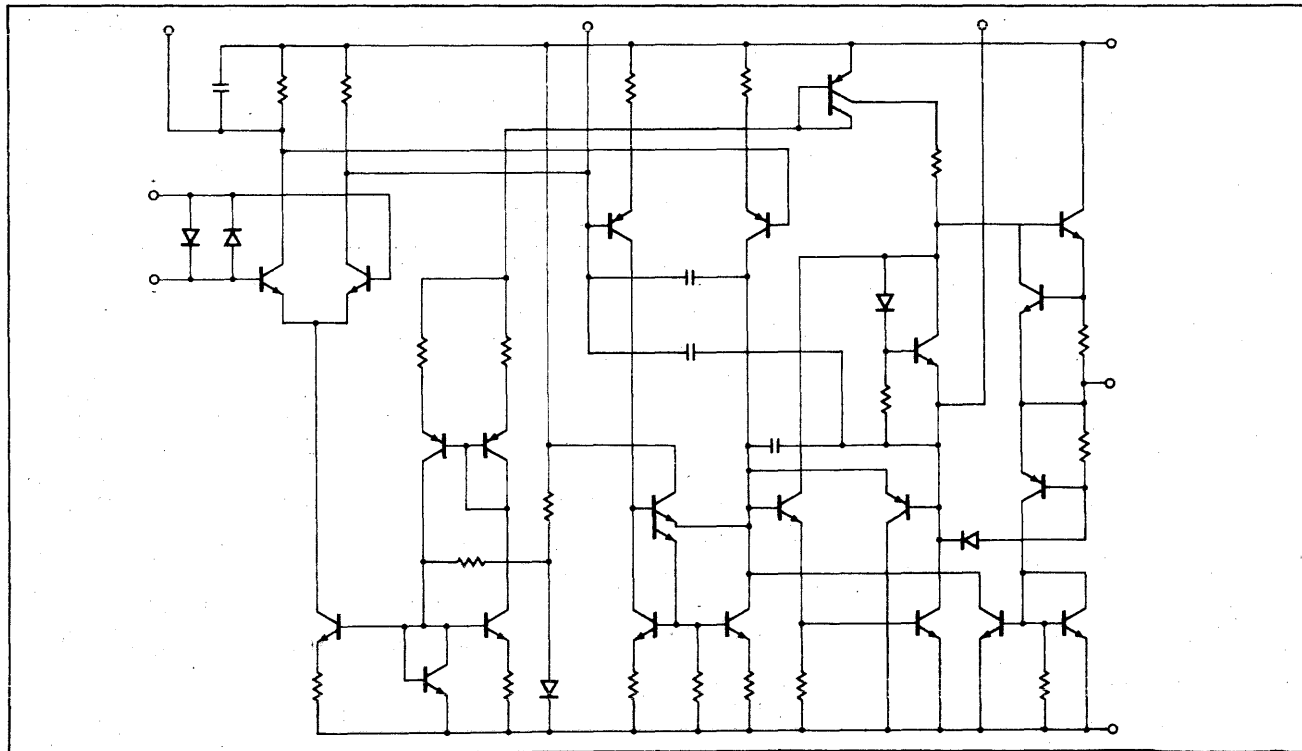
**FEATURES**

- **Small-signal bandwidth: 10MHz**
- **Output drive capability: 600Ω, 10V (rms) at  $V_s = \pm 18V$**
- **Input noise voltage: 4nV/√Hz**
- **DC voltage gain: 100000**
- **AC voltage gain: 6000 at 10kHz**
- **Power bandwidth: 200kHz**
- **Slew-rate: 13V/μs**
- **Large supply voltage range: ±3 to ±20V**

**PIN CONFIGURATIONS**



**EQUIVALENT SCHEMATIC**





**SINGLE AND DUAL LOW NOISE OP AMP NE5533/5533A / NE/SE5534/5534A**

**AC ELECTRICAL CHARACTERISTICS**  $T_A = 25^\circ C, V_S = \pm 15V$  unless otherwise specified.

PARAMETER	TEST CONDITIONS	SE5534 / 5534A			NE5533 / 5533A 5534 / 5534A			UNIT
		Min	Typ	Max	Min	Typ	Max	
R <sub>OUT</sub> Output resistance	$A_V = 30dB$ closed loop $f = 10kHz, R_L = 600\Omega, C_C = 22pF$		0.3			0.3		$\Omega$
Transient response	Voltage follower, $V_{IN} = 50mV$ $R_L = 600\Omega, C_C = 22pF, C_L = 100pF$							
T <sub>R</sub> Rise time			20			20		ns
Overshoot			20			20		%
Transient response	$V_{IN} = 50mv, R_L = 600\Omega$ $C_C = 47pF, C_L = 500pF$							
T <sub>R</sub> Rise time			50			50		ns
Overshoot			35			35		%
AC Gain	$f = 10kHz, C_C = 0$ $f = 10kHz, C_C = 22pF$		6			6		V/mV
Gain bandwidth product	$C_C = 22pF, C_L = 100pF$		2.2			2.2		V/mV
Slew rate	$C_C = 0$ $C_C = 22pF$		10			10		mHz
Power bandwidth	$V_{OUT} = \pm 10V, C_C = 0$ $V_{OUT} = \pm 10V, C_C = 22pF$ $V_{OUT} = \pm 14V, R_L = 600\Omega$ $C_C = 22pF, V_{CC} = \pm 18V$		13			13		V/ $\mu S$
			6			6		V/ $\mu S$
			200			200		kHz
			95			95		kHz
			70			70		kHz

**ELECTRICAL CHARACTERISTICS**  $T_A = 25^\circ C, V_S = \pm 15V$  unless otherwise specified.

PARAMETER	TEST CONDITIONS	5533/5534			5533A/5534A			UNIT
		Min	Typ	Max	Min	Typ	Max	
Input noise voltage	$f_o = 30Hz$ $f_o = 1kHz$		7			5.5	7	nV/ $\sqrt{Hz}$
			4			3.5	4.5	nV/ $\sqrt{Hz}$
Input noise current	$f_o = 30Hz$ $f_o = 1kHz$		2.5			1.5		pA/ $\sqrt{Hz}$
			0.6			0.4		pA/ $\sqrt{Hz}$
Broadband noise figure	$f = 10Hz - 20kHz, R_S = 5k\Omega$					0.9		dB
Channel separation	$f = 1kHz, R_S = 5k\Omega$		110			110		dB

# SINGLE AND DUAL LOW NOISE OP AMP NE5533/5533A / NE/SE5534/5534A

## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
V <sub>S</sub> Supply voltage	± 22	V
V <sub>IN</sub> Input voltage	± V supply	V
V <sub>DIFF</sub> Differential input voltage <sup>1</sup>	± 0.5	V
T <sub>A</sub> Operating temperature range		
SE5534/5534A	- 55 to + 125	°C
NE5533/5533A/5534/5534A	0 to + 70	°C
T <sub>STG</sub> Storage temperature	- 65 to + 150	°C
T <sub>J</sub> Junction temperature	150	°C
P <sub>D</sub> Power dissipation at 25°C <sup>2</sup>	800	mW
5533N, 5534N, 5534FE		
Output short circuit duration <sup>3</sup>	indefinite	
Lead temperature (soldering, 10 sec)	300	°C

NOTES

- Diodes protect the inputs against over-voltage. Therefore, unless current-limiting resistors are used, large currents will flow if the differential input voltage exceeds 0.6V. Maximum current should be limited to ±10mA.
- For operation at elevated temperature, derate packages based on the following junction-to-ambient thermal resistances:
  - 8-pin ceramic (FE) 140°C/W
  - 14-pin ceramic (F) 110°C/W
  - 8-pin plastic (N) 162°C/W
  - 14-pin plastic (N) 150°C/W
- Output may be shorted to ground at V<sub>S</sub> = ±15V, T<sub>A</sub> = 25°C. Temperature and/or supply voltages must be limited to ensure dissipation rating is not exceeded.

## DC ELECTRICAL CHARACTERISTICS T<sub>A</sub> = 25°C, V<sub>S</sub> = ± 15V unless otherwise specified.<sup>1,2</sup>

PARAMETER	TEST CONDITIONS	SE5534/5534A			NE5533/5533A 5534/5534A			UNIT
		Min	Typ	Max	Min	Typ	Max	
V <sub>OS</sub> Offset voltage	Over temperature		0.5	2		0.5	4	mV
ΔV <sub>OS</sub> /ΔT			5	3		5	5	mV/°C
I <sub>OS</sub> Offset current	Over temperature		10	200		20	300	nA
ΔI <sub>OS</sub> /ΔT			200	500		200	400	nA/°C
I <sub>B</sub> Input current	Over temperature		400	800		500	1500	nA
ΔI <sub>B</sub> /ΔT			5	1500		5	2000	nA/°C
I <sub>CC</sub> Supply current Per op amp	Over temperature		4	6.5		4	8	mA
					9		10	mA
V <sub>CM</sub> Common mode input range		± 12	± 13		± 12	± 13		V
CMRR Common mode rejection ratio		80	100		70	100		dB
PSRR Power supply rejection ratio			10	50		10	100	μV/V
A <sub>VOL</sub> Large signal voltage gain	R <sub>L</sub> ≥ 600Ω, V <sub>O</sub> = ± 10V Over temperature	50	100		25	100		V/mV
V <sub>OUT</sub> Output swing 5534 only	R <sub>L</sub> ≥ 600Ω	± 12	± 13		± 12	± 13		V
	Over temperature	± 10	± 12		± 10	± 12		V
	R <sub>L</sub> ≥ 600Ω, V <sub>S</sub> = ± 18V	± 15	± 16		± 15	± 16		V
	R <sub>L</sub> ≥ 2kΩ	± 13	± 13.5		± 13	± 13.5		V
	Over Temperature	± 12	± 12.5		± 12	± 12.5		V
R <sub>IN</sub> Input resistance		50	100		30	100		kΩ
I <sub>SC</sub> Output short circuit current			38			38		mA

NOTES

- For NE5533/5533A/5534/5534A, T<sub>MIN</sub> = 0°C, T<sub>MAX</sub> = 70°C
- For SE5534/5534A, T<sub>MIN</sub> = - 55°C, T<sub>MAX</sub> = + 125°C

LINEAR Signetics

PHASE LOCKED LOOP

NE/SE564

DESCRIPTION

The NE564 is a versatile, high guaranteed frequency Phase Locked Loop designed for operation up to 50MHz. As shown in the block diagram, the NE564 consists of a VCO, limiter, phase comparator, and post detection processor.

APPLICATIONS

- High speed modems
- FSK receivers and transmitters
- Frequency synthesizers
- Signal generators
- Various satcom/TV systems

FEATURES

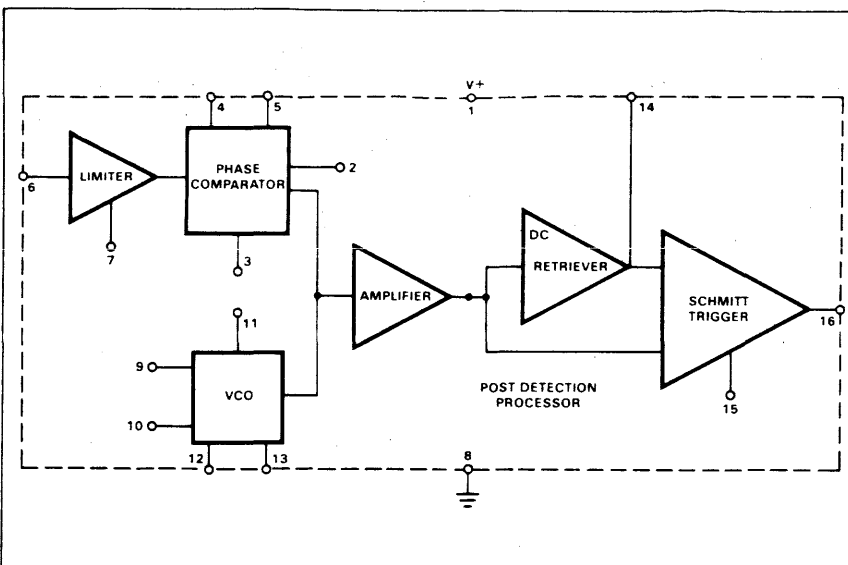
- Operation with single 5V supply
- TTL compatible inputs and outputs
- Guaranteed operation to 50MHz
- External loop gain control
- Reduced carrier feedthrough
- No elaborate filtering needed in FSK applications
- Can be used as a modulator
- Variable loop gain (Externally Controlled)

ABSOLUTE MAXIMUM RATINGS

PARAMETER		RATING	UNIT
V+	Supply voltage		V
	Pin 1	14	
	Pin 10	6	
P <sub>D</sub>	Power dissipation	600	mW
T <sub>A</sub>	Operating temperature	0 to 70	°C
	Operating temperature	-55 to +125	
t <sub>stg</sub>	Storage temperature	-65 to 150	°C

NOTE:  
Operation above 5 volts will require heatsinking of the case.

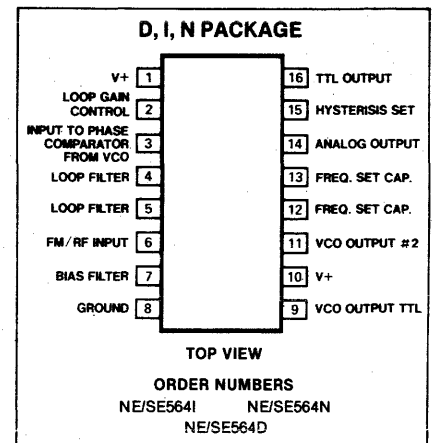
BLOCK DIAGRAM



OUTLINE OF SETUP PROCEDURE

1. Determine operating frequency of the VCO - .  
If  $\div N$  in feedback loop, then  $f_o = N \times f_{in}$ .
2. Calculate value of the VCO frequency set capacitor:  
$$C_o \approx \frac{1}{2500 f_o}$$
3. Set I<sub>2</sub> (current sinking into Pin 2) for  $\approx 200\mu A$ . After operation is obtained, this value may be adjusted for best dynamic behavior.
4. Check VCO output frequency with digital counter at Pin 9 of device (loop open, VCO to  $\phi$  det.). Adjust C<sub>o</sub> trim or frequency adj. Pin 4-5 for exact center frequency if needed.
5. Close loop and inject input signal to Pin 6. Monitor Pin 3 and 6 with two channel scope. Lock should occur with  $\Delta\phi_{3-6}$  equal to 90° (phase error).
6. If pulsed burst or ramp frequency is used for input signal, special loop filter design may be required in place of simple single capacitor filter on Pin 4 and 5.

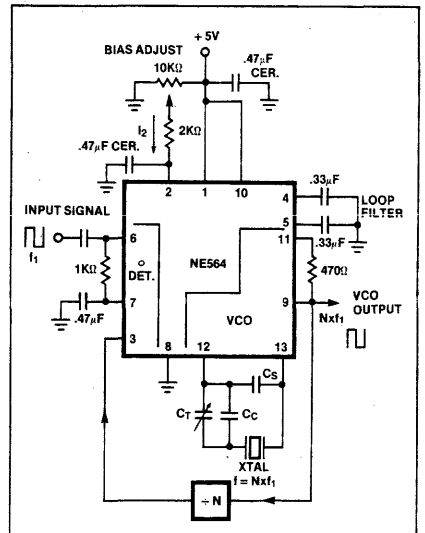
PIN CONFIGURATION



design may be required in place of simple single capacitor filter on Pin 4 and 5. (See PLL application section in Analog Manual.)

7. The input signal to Pin 6 and the VCO feedback signal to Pin 3 must have a duty cycle of 50% for proper operation of the phase detector. Due to the nature of a balanced mixer if signals are not 50% in duty cycle, D.C. offsets will occur in the loop which tend to create an artificial or biased VCO offset.
8. For multiplier circuits where phase jitter is a problem, loop filter capacitors may be increased to a value of 10-50 $\mu F$  on Pin 4, 5. Also careful supply decoupling may be necessary. This includes the counter chain V<sub>CC</sub> lines.

NE564 PHASE LOCKED FREQUENCY MULTIPLIER WITH VCXO



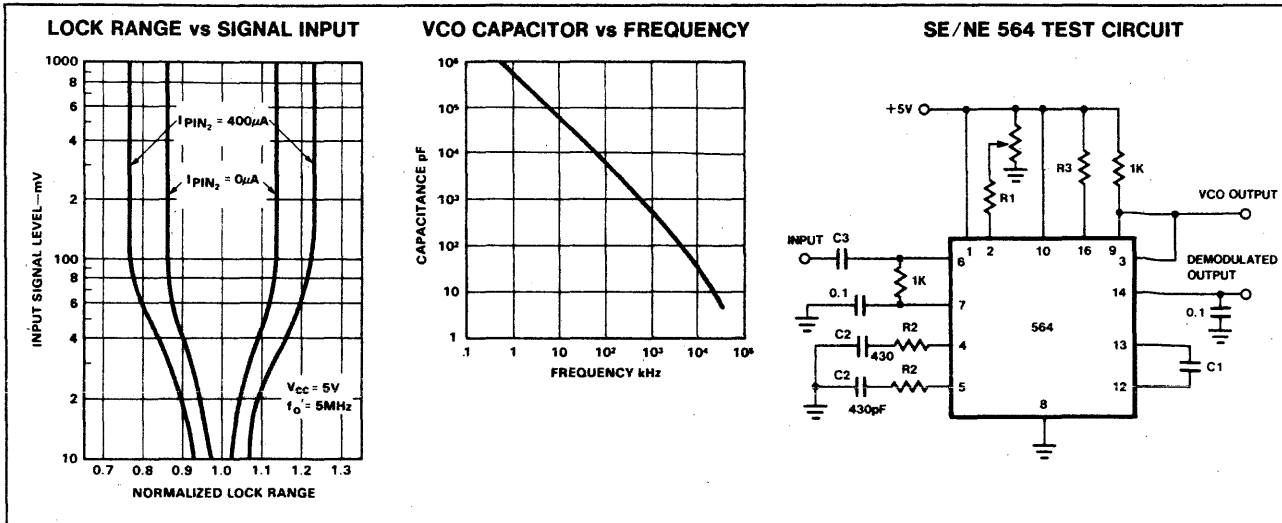
**PHASE LOCKED LOOP**

**NE/SE564**

**ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5V$ ,  $T_A = 25^\circ C$ ,  $f_o = 5MHz$ ,  $I_B = 400\mu A$  unless otherwise specified

PARAMETER	TEST CONDITIONS	SE564			NE564			UNIT
		Min	Typ	Max	Min	Typ	Max	
Maximum VCO frequency	$C_1 = 0$	50	65		45	60		MHz
Lock range	Input $\geq 200mV_{rms}$ $T_A = 25^\circ C$ = $125^\circ C$ = $-55^\circ C$ = $0^\circ C$ = $70^\circ C$	40 20 50	70 30 80		40	70 70 40		% of $f_o$
Capture range	Input $\geq 200mV_{rms}$ , $R_2 = 27\Omega$	20	30		20	30		% of $f_o$
VCO frequency drift with temperature	$f_o = 5MHz$ , $T_A = -55^\circ C$ to $125^\circ C$ = $0^\circ C$ to $70^\circ C$ $f_o = 500KHz$ , $T_A = -55^\circ C$ to $125^\circ C$ = $0^\circ C$ to $70^\circ C$		400 250	1000 500		400 400	1250 850	PPM/ $^\circ C$
VCO free running frequency	$f_o = \frac{1}{25R_C C_1}$ , $C_1 = 80pF$ $R_C = 100\Omega$ "Internal"	4	5	6	3.5	5	7	MHz
VCO frequency change with supply voltage	$V_{CC} = 4.5V$ to $5.5V$		3	8		3	8	% of $f_o$
Demodulated output voltage	Modulation frequency: 1KHz $f_o = 5MHz$ , input deviation: 2% T = $25^\circ C$ 1% T = $25^\circ C$ = $0^\circ C$ = $-55^\circ C$ = $70^\circ C$ = $125^\circ C$	16 8 6 12	28 14 10 16		16 8	28 14 13 15		mVrms mVrms mVrms mVrms mVrms
Distortion	Deviation: 1% to 8%		1			1		%
Signal to noise ratio	Std. condition, 1% to 10% dev.		40			40		dB
AM rejection	Std. condition, 30% AM		35			35		dB
Demodulated Output at operating voltage	Modulation frequency: 1KHz $f_o = 5MHz$ , input deviation: 1% $V_{CC} = 4.5V$ $V_{CC} = 5.5V$	7 8	12 14		7 8	12 14		mVrms mVrms
Supply current	$V_{CC} = 5V$ $I_1, I_{10}$		45	60		45	60	mA
Output "1" output leakage current	$V_{OUT} = 5V$ , Pin 16, 9		1	20		1	20	$\mu A$
Output "0" output voltage	$I_{OUT} = 2mA$ , Pin 16, 9 $I_{OUT} = 6mA$ , Pin 16, 9		0.3 0.4	0.6 0.8		0.3 0.4	0.6 0.8	V V

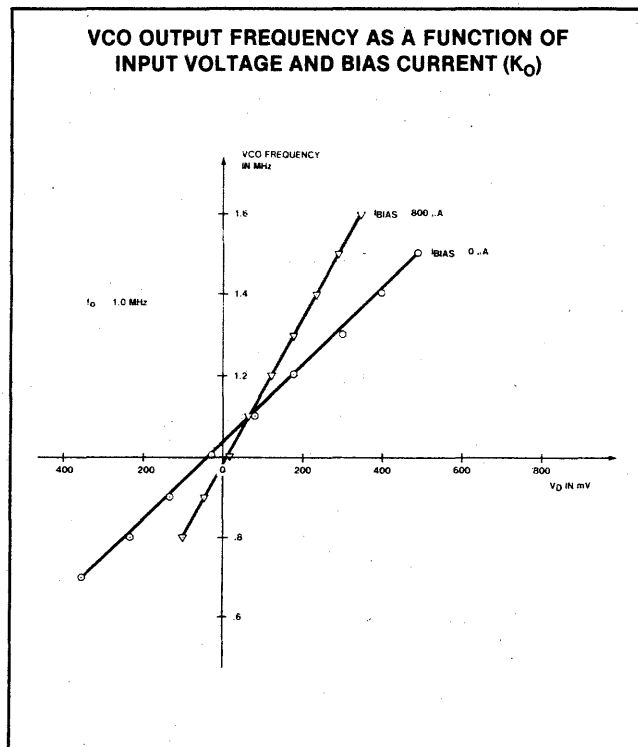
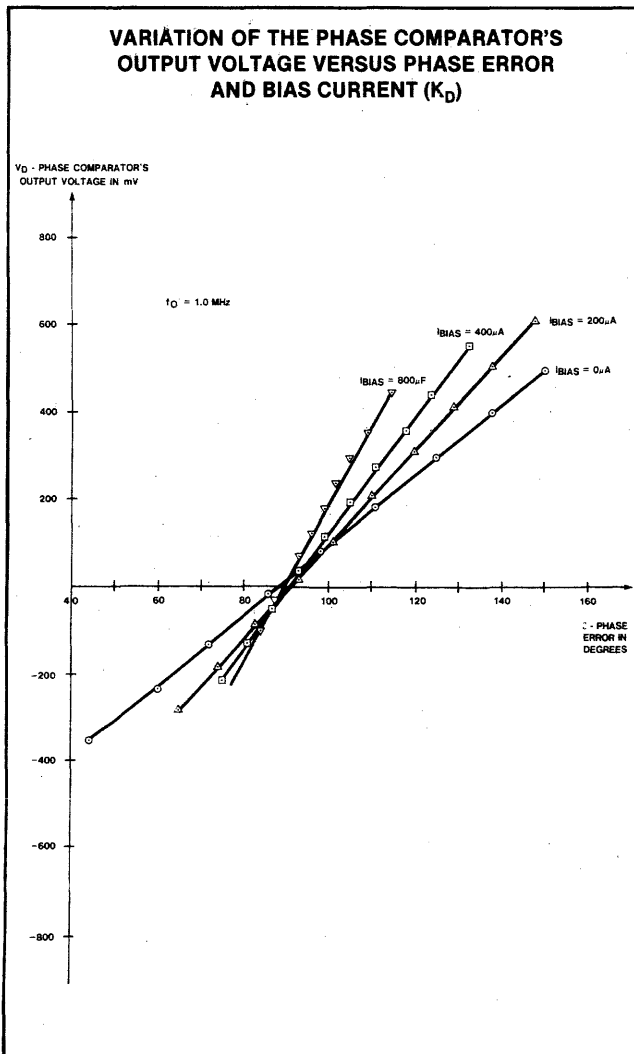
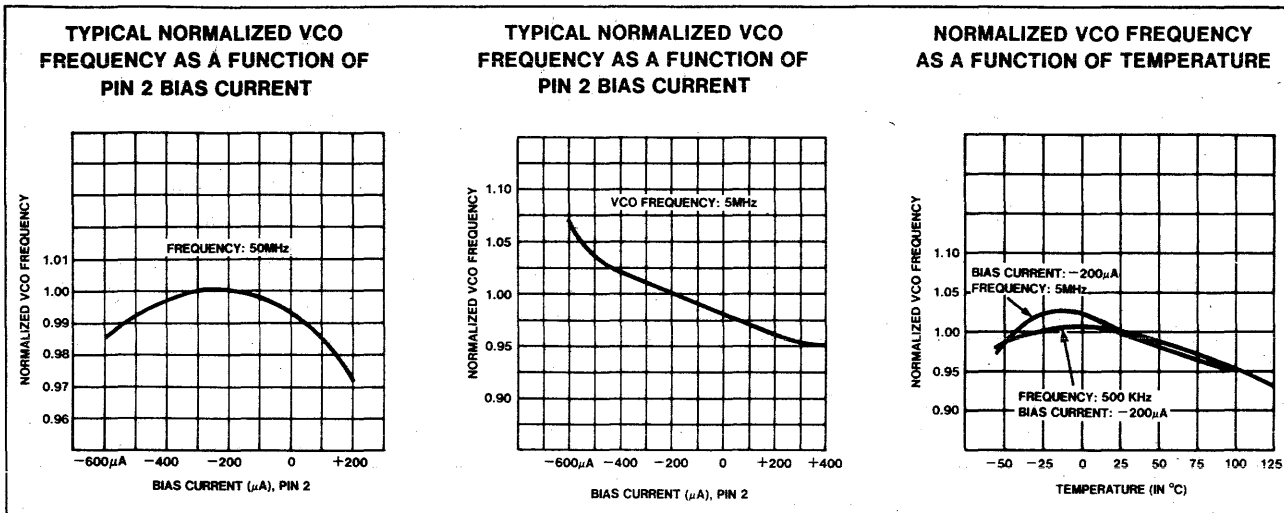
**TYPICAL PERFORMANCE CHARACTERISTICS**



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PHASE LOCKED LOOP

NE/SE564



LINEAR

Signetics

# SWITCHED-MODE POWER SUPPLY CONTROL CIRCUIT

NE/SE5560

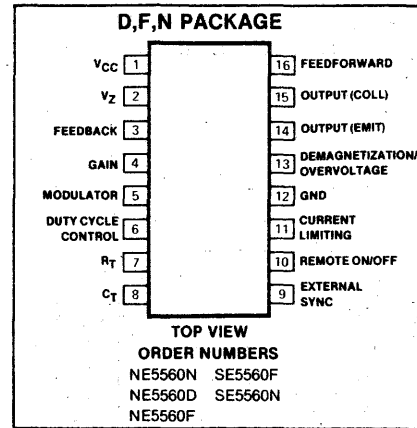
## DESCRIPTION

The NE/SE5560 is a control circuit for use in switched mode power supplies. This single monolithic chip incorporates all the control and housekeeping (protection) functions required in switched mode power supplies, including an internal temperature compensated reference source, internal Zener reference, sawtooth generator, pulse width modulator, output stage and various protection circuits.

## FEATURES

- Stabilized power supply
- Temperature compensated reference source
- Sawtooth generator
- Pulse width modulator
- Remote on/off switching
- Current limiting
- Low supply voltage protection
- Loop fault protection
- Demagnetization/overvoltage protection
- Maximum duty cycle clamp
- Feed forward control
- External synchronization

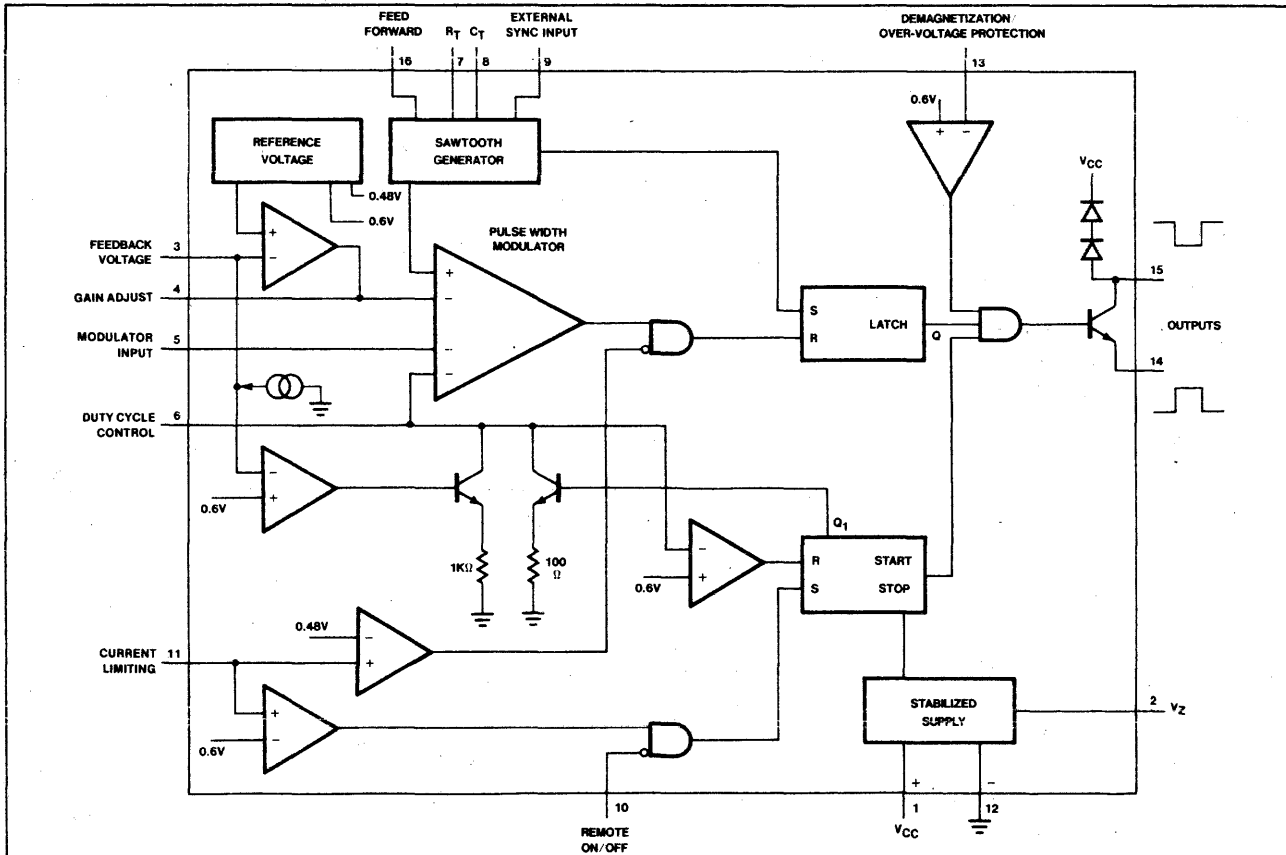
## PIN CONFIGURATION



## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Supply Voltage sourced	+18	V
Current sourced	30	mA
Output transistor Output current	40	mA
Collector voltage (Pin 15)	+18	V
Max. emitter voltage (Pin 14)	+5	V
Operating temperature (ambient)		
SE5560	-55 to +125	°C
NE5560	0 to 70	°C
Storage temperature range	-65 to +150	°C

## BLOCK DIAGRAM



LINEAR Signetics

## SWITCHED-MODE POWER SUPPLY CONTROL CIRCUIT

NE/SE5560

DC ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 12\text{V}$  unless otherwise specified)

PARAMETER	TEST CONDITIONS	SE5560			NE5560			UNIT
		Min	Typ	Max	Min	Typ	Max	
<b>Reference Sections</b>								
Internal reference voltage ( $V_{ref}$ )	25°C	3.69	3.72	3.81	3.57	3.72	3.95	V
Temperature coefficient of $V_{ref}$	Over temperature	3.65		3.85	3.53		4.00	V
Internal Zener reference ( $V_Z$ )	$I_L = -7\text{ mA}$	7.8	-100	9.0	7.8	-100	9.0	ppm/°C
Temperature coefficient of $V_Z$			8.4	200		-200		V
								ppm/°C
<b>Oscillator Section</b>								
Frequency range	Over temperature	50		100k	50		100k	Hz
Initial accuracy oscillator	$R = 5\text{ k}\Omega$		5			5		%
Duty cycle range	$f_o = 20\text{ kHz}$	0		98	0		98	%
<b>Modulator</b>								
Modulation input current	Voltage at Pin 5 = 2V Over temperature		0.2	20		0.2	20	$\mu\text{A}$
<b>Housekeeping Function</b>								
Pin 6, input current	at 2V Over temperature		0.2	20		0.2	20	$\mu\text{A}$
Pin 6, duty cycle limit control	(for 50% maximum duty cycle) 15 kHz to 50 kHz/ 41% of $V_Z$	40	50	60	40	50	60	% of duty cycle
Pin 1, low supply voltage protection thresholds		8	9.0	10.5	8	9.0	10.5	V
Pin 3, feedback loop protection trip threshold		400	600	720	400	600	720	mV
Pin 3, pull up current	at 2V Over temperature	-7	-15	-35	-7	-15	-35	$\mu\text{A}$
Pin 13, demagnetization/over voltage protection trip on threshold		470	600	720	470	600	720	mV
Pin 13, input current	at 0.25V 25°C		0.6	10		0.6	10	$\mu\text{A}$
Pin 16, feed forward duty cycle control	Over temperature Voltage at Pin 16 = $2V_Z$	30	40	50	30	40	50	% original duty cycle
*Pin 16, feed forward input current	at 16V, $V_{CC} = 18\text{V}$ 25°C		0.2	5		0.2	5	$\mu\text{A}$
	Over temperature			10			10	$\mu\text{A}$
<b>External Synchronization</b>								
Pin 9 off		0		0.8	0		0.8	V
Pin 9 on		2		$V_Z$	2		$V_Z$	V
Pin 9 sink current	Voltage at Pin 9 = 0V, 25°C Over temperature		-65	-100		-65	-125	$\mu\text{A}$
				-125			-125	$\mu\text{A}$
<b>Remote</b>								
Pin 10 off		0	0.8		0	0.8		V
Pin 10 on		2	$V_Z$		2	$V_Z$		V
Pin 10 sink current	at 0V 25°C		-85	-100		-85	-125	$\mu\text{A}$
	Over temperature			-125			-125	$\mu\text{A}$
<b>Current Limiting</b>								
Pin 11, $I_{IN}$	Voltage at Pin 11 = 250 mV, 25°C		-2	-10		-2	-10	$\mu\text{A}$
Single pulse inhibit delay	Over temperature Inhibit delay time for 20% overdrive at 40 mA $I_{OUT}$		0.7	0.8		0.7	0.8	$\mu\text{s}$
Trip Levels: Shut down, slow start		0.560	0.600	0.700	0.560	0.600	0.700	V
Current limit		0.400	0.480	0.500	0.400	0.480	0.500	V
<b>Error Amplifier</b>								
Output voltage swing ( $V_{OH}$ )		6.2		9.5	6.2		9.5	V
Output voltage swing ( $V_{OL}$ )				0.7			0.7	V
Open loop gain		54	60		54	60		dB
Feedback resistor		10k			10k			$\Omega$
Small signal bandwidth			3			3		MHZ

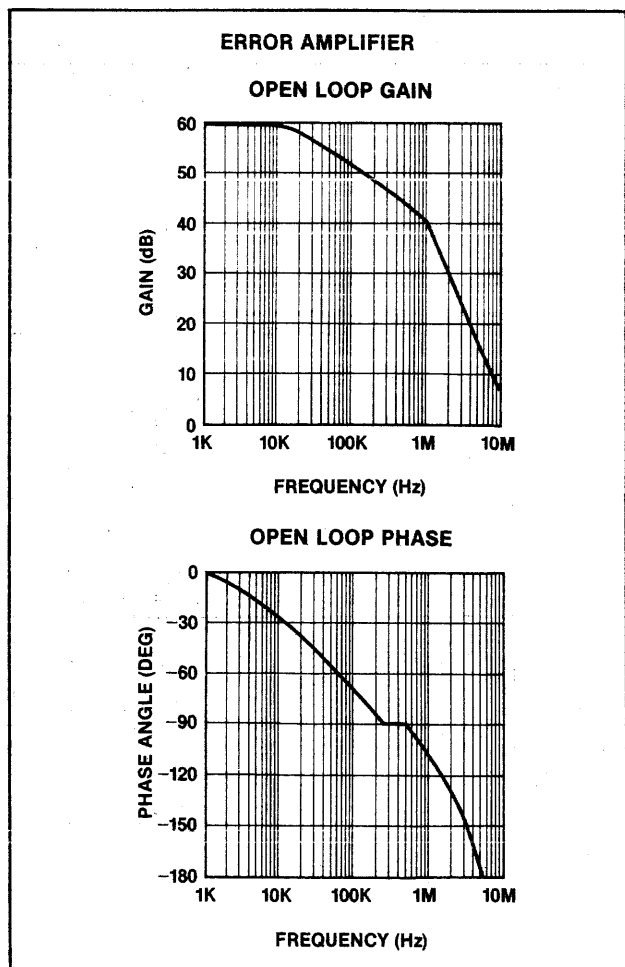
**SWITCHED-MODE POWER SUPPLY CONTROL CIRCUIT**

**NE/SE5560**

**DC ELECTRICAL CHARACTERISTICS** (Continued)

PARAMETER	TEST CONDITIONS	SE5560			NE5560			UNIT
		Min	Typ	Max	Min	Typ	Max	
<b>Output Stage</b> V <sub>CE(SAT)</sub> I <sub>C</sub> = 40mA Output current (pin 15) Max emitter voltage (pin 14)		40 5		0.5	40 5		0.5	V mA V
<b>Supply Voltage/Current</b> I <sub>CC</sub>	I <sub>Z</sub> = 0, voltage fed, V <sub>CC</sub> = 12V, 25°C Over temp. I <sub>CC</sub> = 10mA, current feed I <sub>CC</sub> = 30mA, current feed			10 15			10 15	mA mA
V <sub>CC</sub>		20		23	19		24	V
V <sub>CC</sub>		20		30	20		30	V

**TYPICAL PERFORMANCE CHARACTERISTICS**



**MAXIMUM PIN VOLTAGES**

NE5560	
FUNCTION	MAXIMUM VOLTAGE
1. V <sub>CC</sub>	+ 18 volts
2. V <sub>Z</sub>	Do not force (28.4V)
3. Feedback	V <sub>Z</sub>
4. Gain	
5. Modulator	V <sub>Z</sub>
6. Duty Cycle Control	V <sub>Z</sub>
7. R <sub>T</sub>	Current force mode
8. C <sub>T</sub>	
9. External Sync	V <sub>Z</sub>
10. Remote On/Off	V <sub>Z</sub>
11. Current Limiting	V <sub>CC</sub>
12. GND	GND
13. Demagnetization/Overvoltage	V <sub>CC</sub>
14. Output (Emit)	V <sub>Z</sub>
15. Output (Collector)	V <sub>CC</sub> + 2V <sub>be</sub>
16. Feed forward	V <sub>CC</sub>

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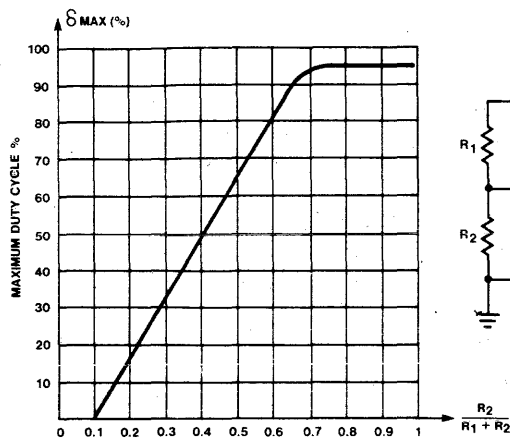


SWITCHED-MODE POWER SUPPLY CONTROL CIRCUIT

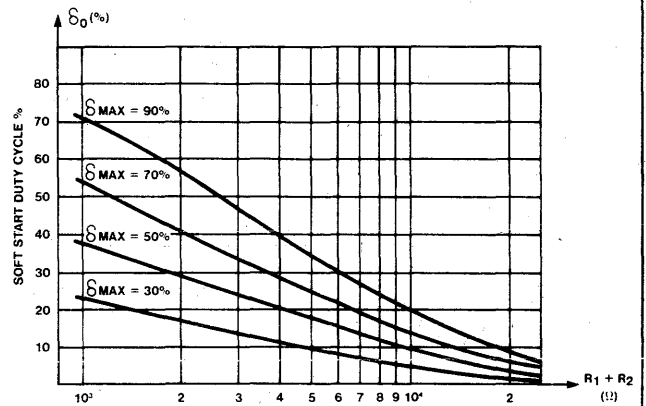
NE/SE5560

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

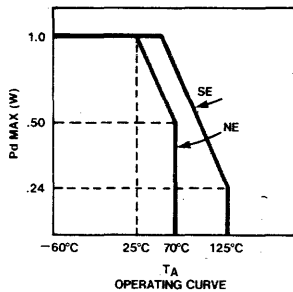
GRAPH FOR DETERMINING  $\delta$  MAX



SOFT-START MIN DUTY CYCLE vs  $R_1 + R_2$

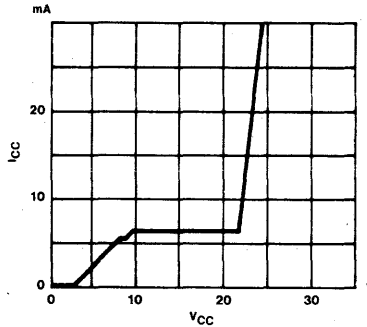


POWER DERATING CURVE

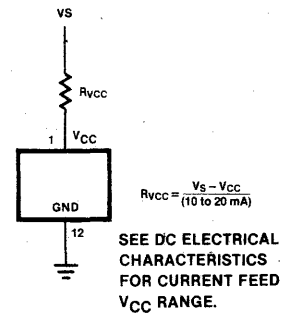


$$Pd = V_{CC} I_{CC} + (V_{CC} - V_Z) I_Z + [(V_{15} - V_{14})_{15} \times \text{DUTY CYCLE}]$$

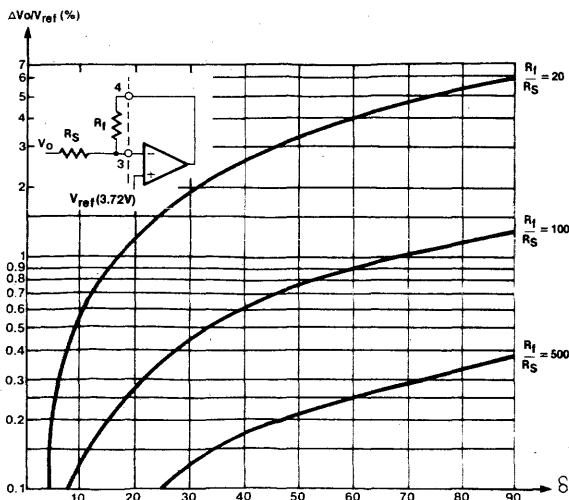
NE5560 VOLTAGE/CURRENT FED SUPPLY CHARACTERISTICS



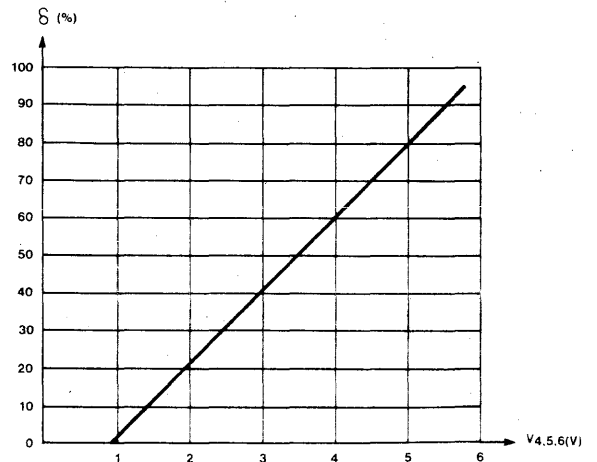
CURRENT FED DROPPING RESISTOR



REGULATION vs ERROR AMP CLOSED LOOP GAIN



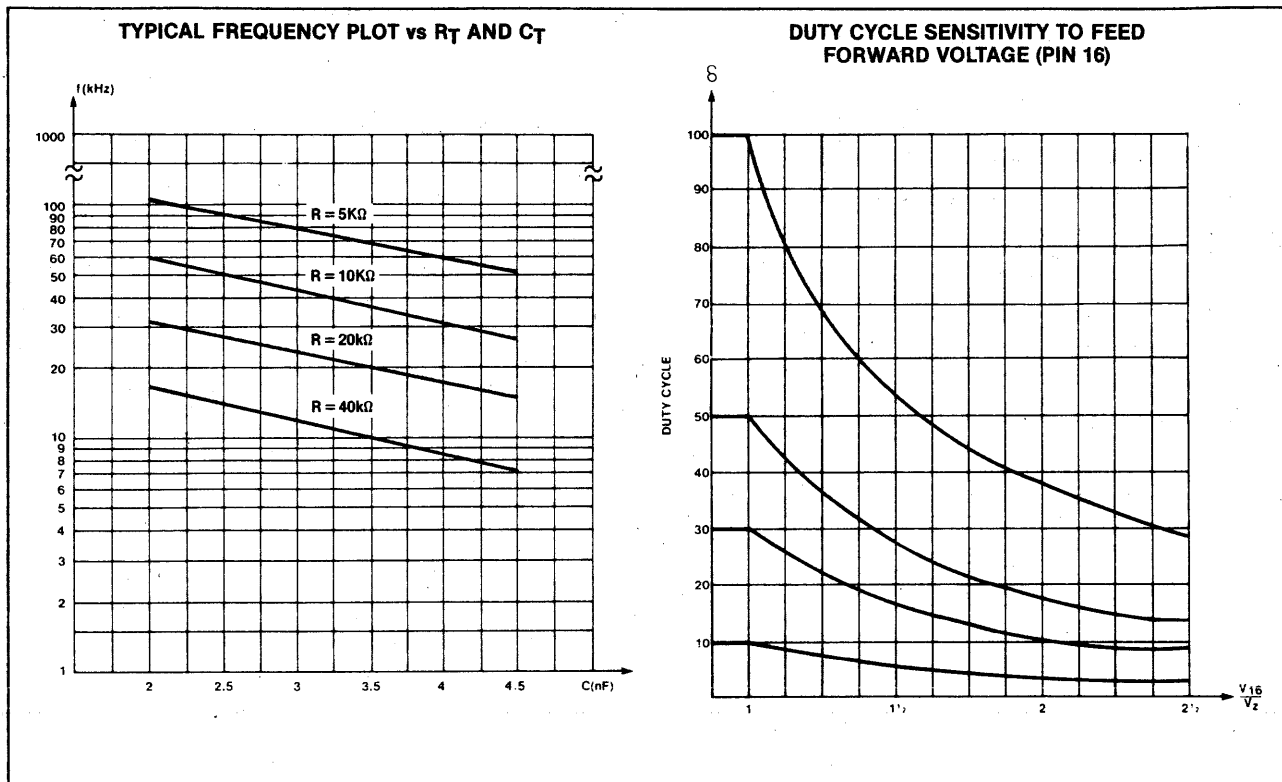
TRANSFER CURVE OF PULSE WIDTH MODULATOR DUTY CYCLE vs INPUT VOLTAGE



SWITCHED-MODE POWER SUPPLY CONTROL CIRCUIT

NE/SE5560

TYPICAL PERFORMANCE CHARACTERISTICS (continued)



THEORY OF OPERATION

- The following functions are incorporated:
- A temperature compensated reference source.
- An error amplifier with pin 3 as input. The output is connected to pin 4 so that the gain is adjustable with external resistors.
- A sawtooth generator with a TTL-compatible synchronization input (pins 7,8,9).
- A pulse-width modulator with a duty-cycle range from 0 to 95%.

(The PWM has two additional inputs:

Pin 6 can be used for a precise setting of  $\delta$  max.

Pin 5 gives a direct access to the modulator, allowing for real constant current operation:)

- A gate at the output of the PWM provides a simple dynamic current limit.
- A latch that is set by the flyback of the sawtooth and reset by the output pulse of the above-mentioned gate prohibits double pulsing.
- Another latch functions as a start-stop circuit; it provides a fast switch-off and a slow start.
- A current protection circuit that operates via the start-stop circuit. This is a combined function with the current

limit circuit, therefore pin 11 has two trip-on levels; the lower one for cycle-by-cycle current limiting, the upper one for current protection by means of switch-off and slow-start.

- A TTL-compatible remote on/off input at pin 10, also operating via the start-stop circuit.
- An inhibit input at pin 13. The output pulse can be inhibited immediately.
- An output gate that is commanded by the latches and the inhibit circuit.
- An output transistor of which both the collector (pin 15) and the emitter (pin 14) are externally available. This allows for normal or inverse output pulses.
- A power supply that can be either voltage or current driven (pins 1 and 12). The internally generated stabilized output voltage  $V_Z$  is connected to pin 2.
- A special function is the so-called feed-forward at pin 16. The amplitude of the sawtooth generator is modulated in such a way that the duty cycle becomes inversely proportional to the voltage on this pin:  $\delta \sim 1/V_{16}$
- Loop fault protection circuits assure that the duty-cycle is reduced to zero or a low value for open or short-circuited feedback loops.

Stabilized Power Supply (Pins 1, 2, 12)

The power supply of the NE5560 is of the well known series regulation type and provides a stabilized output voltage of typically 8.5 volts.

This voltage  $V_Z$  is also present at pin 2 and can be used for precise setting of  $\delta$  max. and to supply external circuitry. Its maximum current capability is 5mA.

The circuit can be fed directly from a DC voltage source between 10.5V and 18V or can be current driven via a limiting resistor. In the latter case, internal pinch-off resistors will limit the maximum supply voltage; typical 23V for 10mA and maximum 30V for 30mA.

The low supply voltage protection is active when  $V(1-12)$  is below 10.5V and inhibits the output pulse (no hysteresis).

When the supply voltage surpasses the 10.5V level, the IC starts delivering output pulses via the slow-start function.

The current consumption at 12V is less than 10mA, provided that no current is drawn from  $V_Z$  and  $R(7-12) \geq 20k\Omega$ .

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**SWITCHED MODE POWER SUPPLY CONTROLLER**

**NE/SE5561**

**DESCRIPTION**

The NE5561/SE5561 is a control circuit for use in switched mode power supplies. It contains an internal temperature compensated supply, PWM, sawtooth oscillator, over-current sense latch, and output stage. The device is intended for low cost SMPS applications where extensive housekeeping functions are not required.

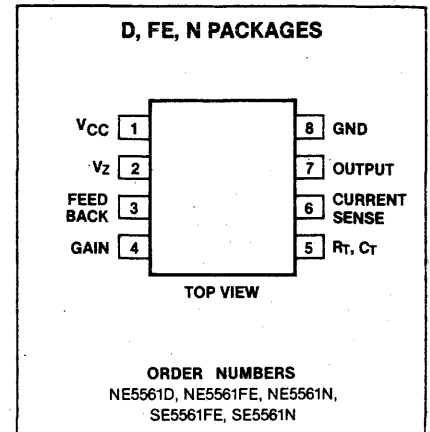
**FEATURES**

- Micro-miniature (D) package
- Pulse-width modulator
- Current limiting (cycle by cycle)
- Sawtooth generator
- Stabilized power supply
- Double pulse protection
- Internal temperature compensated reference

**APPLICATIONS**

- Switch mode power supplies
- D/C motor controller inverter
- DC/DC converter

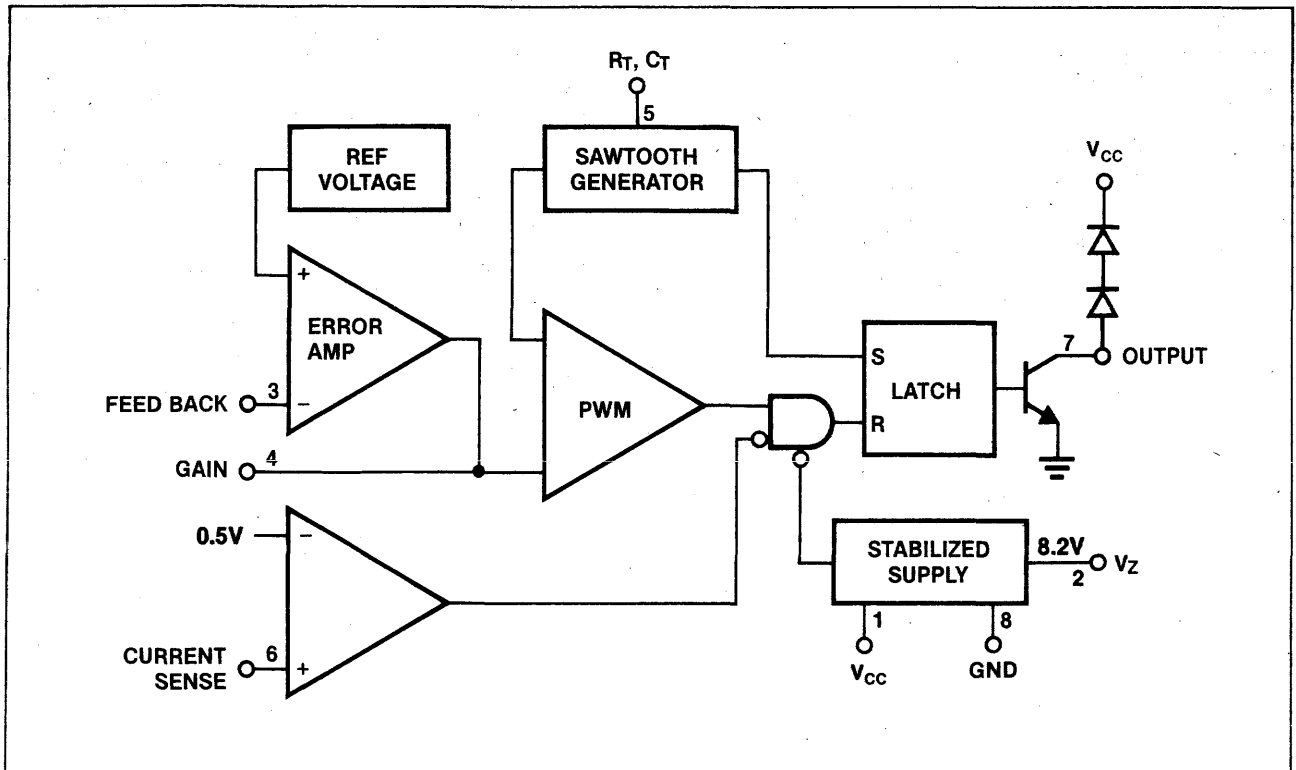
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
Supply voltage, $V_{CC}$	18	V
Output Current	40	mA
Output duty cycle	98	%
Max total power dissipation	0.75	W
Operating temperature range		°C
NE5561	0 to 70	°C
SE5561	-55 to +125	°C

**BLOCK DIAGRAM**



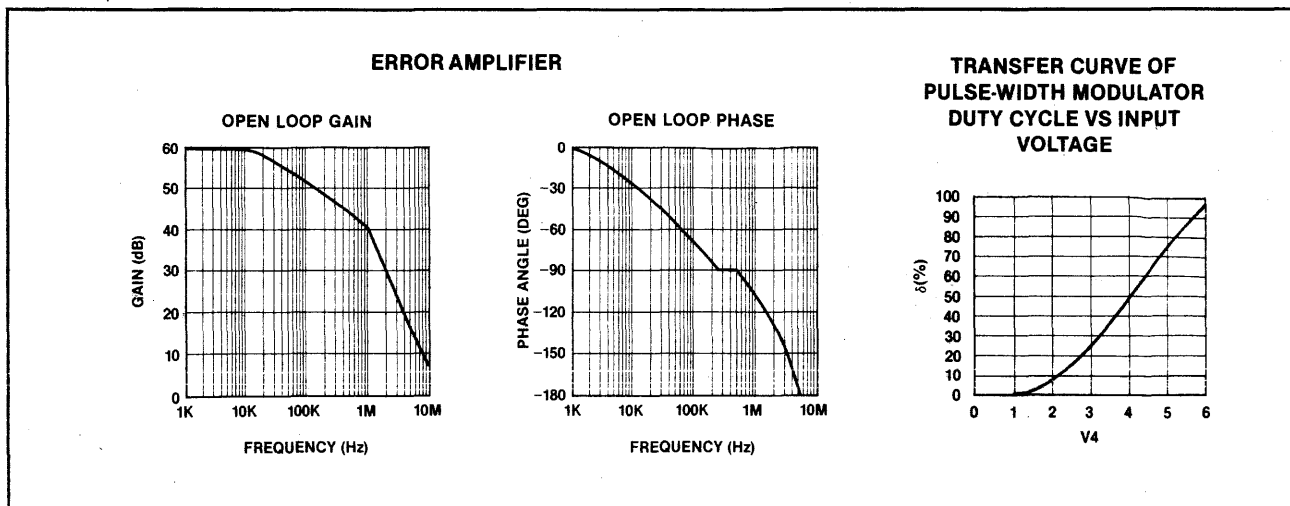
**SWITCHED MODE POWER SUPPLY CONTROLLER**

**NE/SE5561**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 12V$ ,  $T_A = 25^\circ C$  unless otherwise specified.

SYMBOL AND PARAMETER	TEST CONDITIONS	SE5561			NE5561			UNIT	
		Min	Typ	Max	Min	Typ	Max		
<b>Reference Section</b> $V_{REF}$ , Internal ref voltage	$T_A = 25^\circ C$	3.69	3.75	3.84	3.57	3.75	3.96	V	
	Over temp.	3.65		3.88	3.55		3.98	V	
$V_Z$ , internal zener ref	$I_L = 7mA$	7.8	8.2	8.8	7.8	8.2	8.8	V	
Temp coefficient of $V_{REF}$			$\pm 100$			$\pm 100$		ppm/ $^\circ C$	
Temp coefficient of $V_Z$			$\pm 200$			$\pm 200$		ppm/ $^\circ C$	
<b>Oscillator Section</b> Frequency range	Over temp.	50		100k	50		100k	Hz	
		Initial accuracy		12			12	%	
		Duty cycle range	$f_o = 20kHz$	0		98	0		98
<b>Current Limiting</b> ( $I_{IN}$ )	Pin 6 = 250mV	$T_A = 25^\circ C$		-2	-10		-2	-10	$\mu A$
		Over temp.			-20			-20	$\mu A$
Single pulse inhibit delay	Inhibit delay time for 20% overdrive at	$I_{OUT} = 20mA$		0.88	1.10		0.88	1.10	$\mu S$
		$I_{OUT} = 40mA$		0.7	0.8		0.7	0.8	$\mu S$
Current limit trip level		.400	.500	.600	.400	.500	.600	V	
<b>Error Amplifier</b> Open loop gain			60			60		dB	
	Feedback resistor	10k			10k			$\Omega$	
	Small signal bandwidth		3			3		MHz	
	Output voltage swing ( $V_{OH}$ )		6.2			6.2		V	
	Output voltage swing ( $V_{OL}$ )				0.7			0.7	V
	<b>Output Stage</b> Output current	Over temp.	20			20			mA
$V_{ce Sat}$	$I_C = 20mA$ , over temp.			0.4			0.4	V	
<b>Supply Voltage/Current</b>	$I_Z = 0$ , voltage fed	$T_A = 25^\circ C$			10.0			10.0	mA
		Over temp.			13.0			13.0	mA
	$V_{CC}$	$I_S = 10mA$ , current fed	20.0	21.0	22.0	19.0	21.0	24.0	V

**TYPICAL PERFORMANCE CHARACTERISTICS**

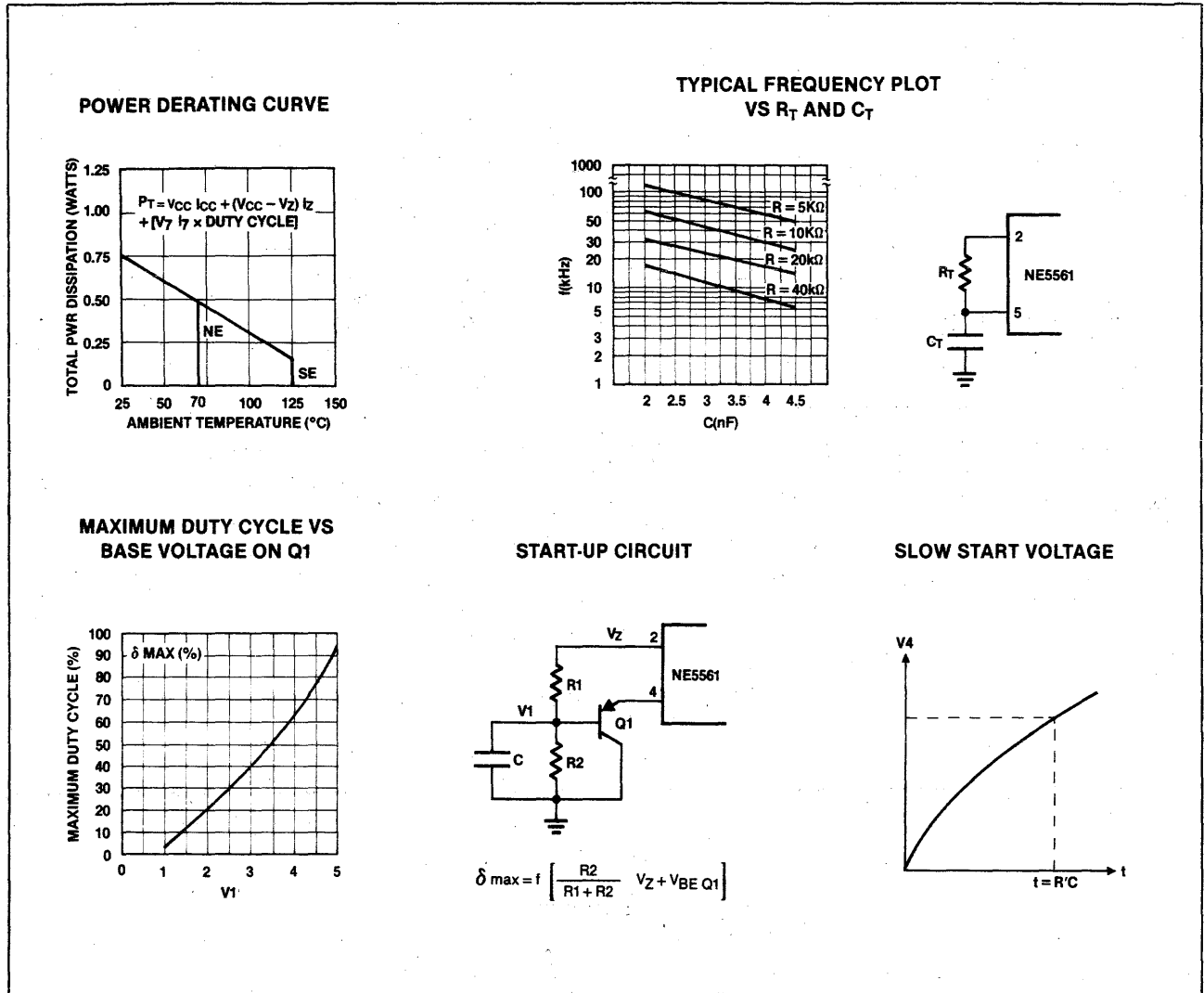


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SWITCHED MODE POWER SUPPLY CONTROLLER

NE/SE5561

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)



NE5561 Start-Up

The start-up, or initial turn on, of this device requires some degree of external protective duty cycle limiting to prevent the duty cycle from initially going to the extreme maximum ( $\delta > 90\%$ ). Either over-current limit or slow start circuitry must be employed to limit duty cycle to a safe value during start-up. Both may be used if desired.

To implement slow-start, the start-up circuit can be used. The divider R1 and R2 sets a voltage, buffered by Q1, such that the output of the error amplifier is clamped to a maximum output voltage, thereby limiting the maximum duty cycle. The addition of capacitor C will cause this voltage to ramp up slowly when power is applied, causing the duty cycle to ramp up simultaneously.

Over-current limit may be used also. To limit duty cycle in this mode, the switch current is monitored at pin 6 and the output of the 5561 is disabled on a cycle by cycle basis when current reaches the programmed limit. With current limit control of slow-start, the duty cycle is limited to that value just allowing maximum switch current to flow. (Approximately 0.50V measured at pin 6.)

APPLICATIONS

5V, 0.5A Buck Regulator Operates from 15V

The converter design shows how simple it is to derive a TTL supply from a system supply of 15V (see Figure 1). The NE5561 drives a 2N4920 PNP transistor directly to provide switching current to the inductor.

Overall line regulation is excellent and covers a range of 12V to 18V with minimal change (< 10 mV) in the output operating at full load.

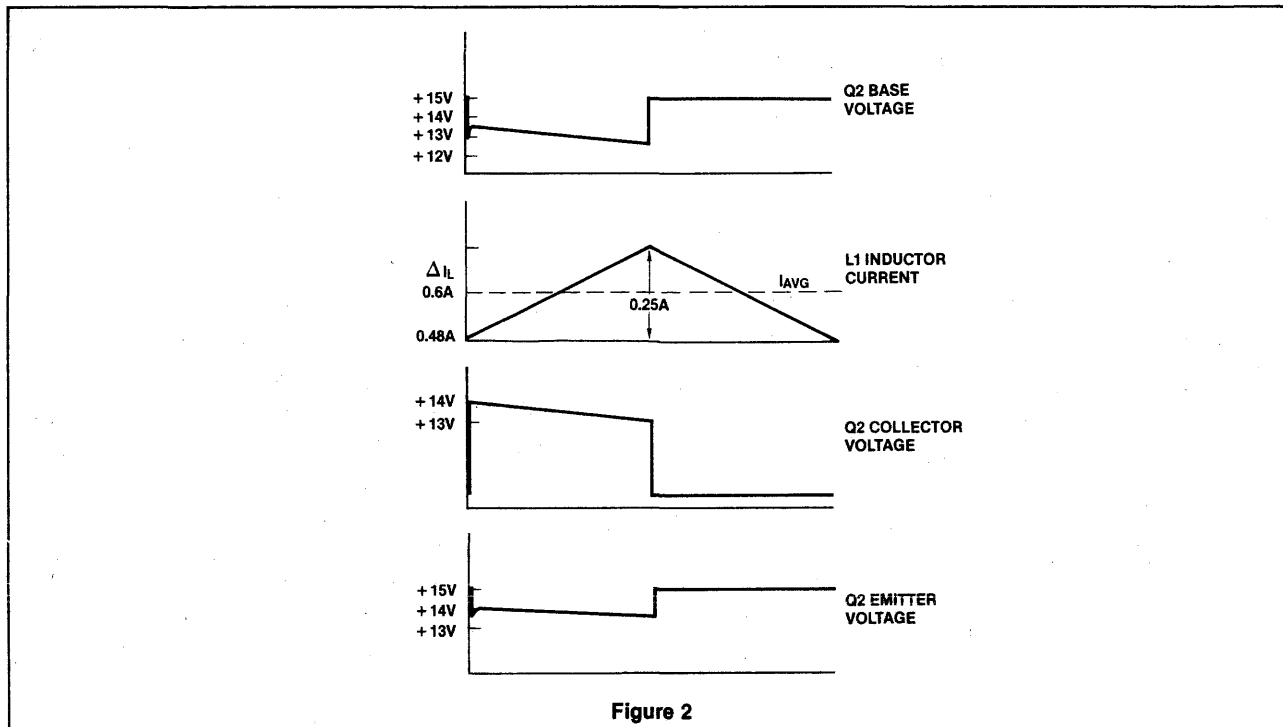
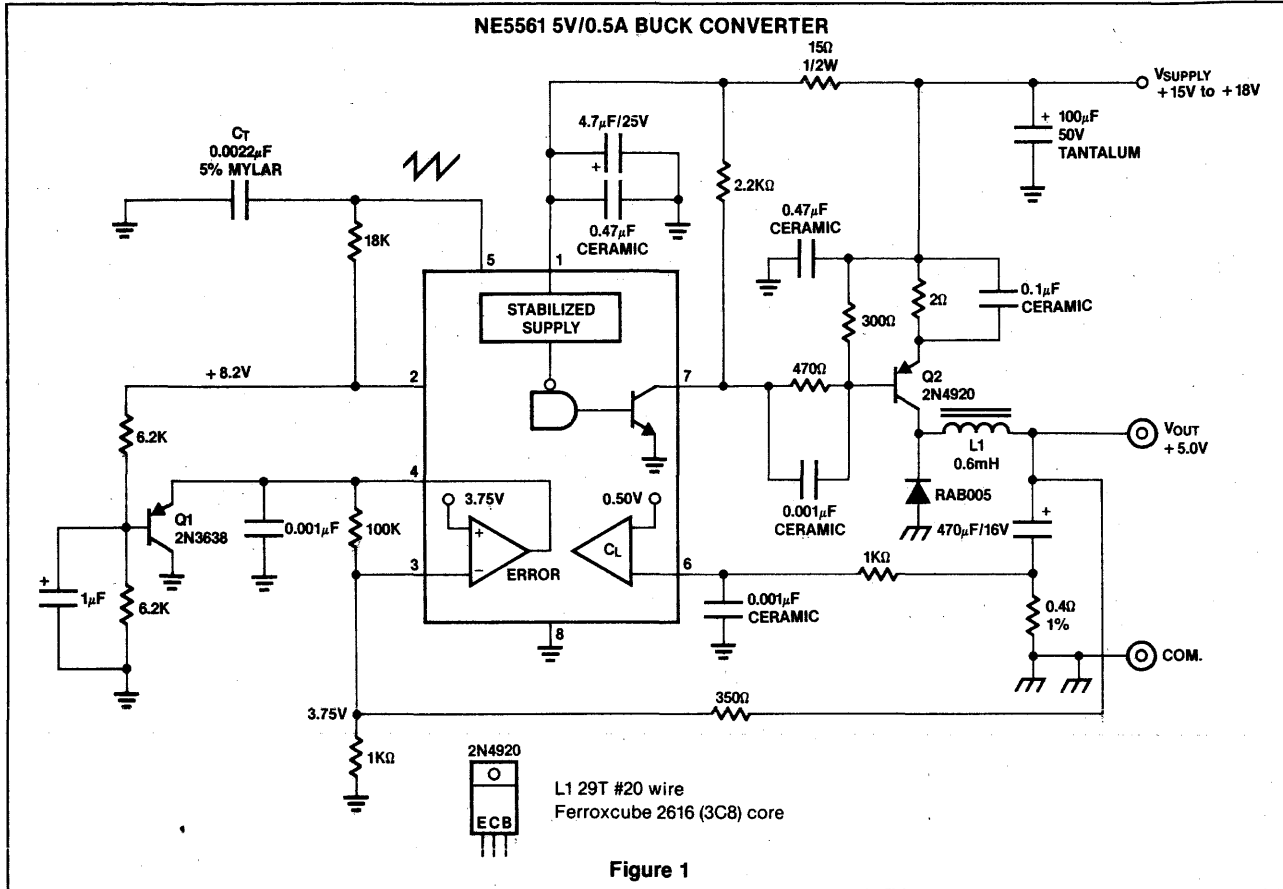
As with all NE5561 circuits, the auxiliary slow start and  $\delta_{max}$  circuit is required, as evidenced by Q1. The  $\delta_{max}$  limit may be calculated by using the relationship (Figure 5a, b).

$$\frac{R2}{R1 + R2} (8.2V) = V\delta_{(max)}$$

The maximum duty cycle is then determined from the pulse-width modulator transfer graph, and R1, R2 are defined from the desired conditions.

SWITCHED MODE POWER SUPPLY CONTROLLER

NE/SE5561



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**REGULATING PULSE WIDTH MODULATOR**

**SG3524**

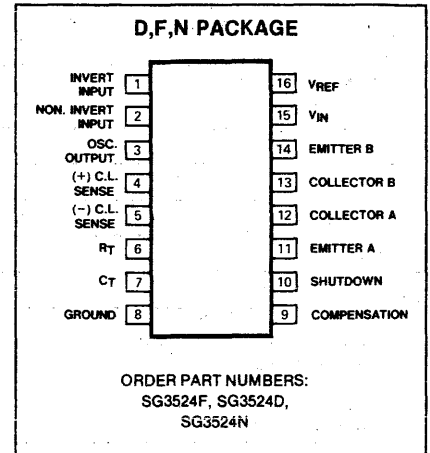
**DESCRIPTION**

This monolithic integrated circuit contains all the control circuitry for a regulating power supply inverter or switching regulator. Included in a 16-pin dual-in-line package is the voltage reference, error-amplifier, oscillator, pulse width modulator, pulse steering flip-flop, dual alternating output switches and current limiting and shut-down circuitry. This device can be used for switching regulators of either polarity, transformer coupled DC to DC converters, transformerless voltage doublers and polarity converters, as well as other power control applications. The SG3524 is designed for commercial applications of 0°C to +70°C.

**FEATURES**

- Complete PWM power control circuitry
- Single ended or push-pull outputs
- Line and load regulation of 0.2%
- 1% maximum temperature variation
- Total supply current is less than 10mA
- Operation beyond 100kHz

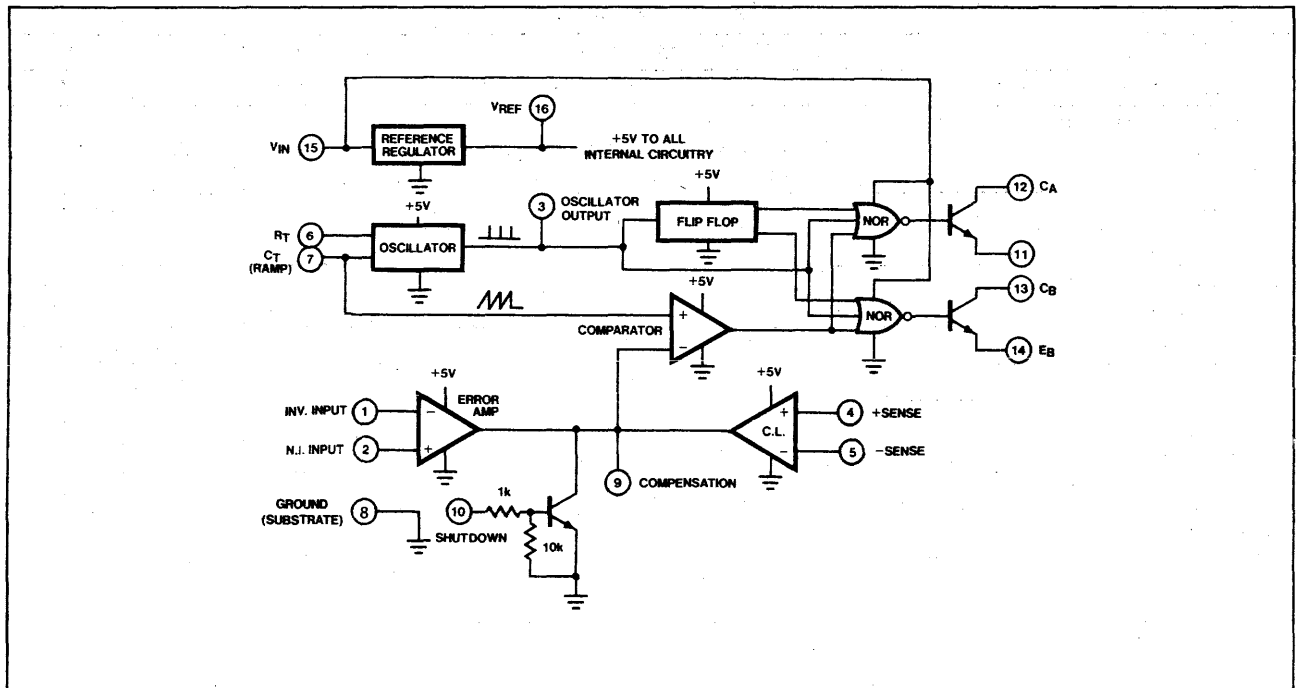
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
Input voltage	40	V
Output current (each output)	100	mA
Reference output current	50	mA
Oscillator charging current	5	mA
Power dissipation		
Package limitation	1000	mW
Derate above 25°C	8	mW/°C
Operating temperature range	0 to +70	°C
Storage temperature range	-65 to +150	°C

**BLOCK DIAGRAM**



**REGULATING PULSE WIDTH MODULATOR**

**SG3524**

**DC ELECTRICAL CHARACTERISTICS** ( $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$ ,  $V_{ON} = 20\text{V}$ , and  $f = 20\text{kHz}$  unless otherwise specified.)

PARAMETER	TEST CONDITIONS	SG3524			UNIT
		Min	Typ	Max	
<b>Reference Section</b>					
Output voltage		4.6	5.0	5.4	V
Line regulation	$V_{IN} = 8$ to $40\text{V}$		10	30	mV
Load regulation	$I_L = 0$ to $20\text{mA}$		20	50	mV
Ripple rejection	$f = 120\text{Hz}$ , $T_A = 25^\circ\text{C}$		66		dB
Short circuit current limit	$V_{REF} = 0$ , $T_A = 25^\circ\text{C}$		100		mA
Temperature stability	Over operating temperature range		0.3	1	%
Long term stability	$T_A = 25^\circ\text{C}$		20		mV/kHr
<b>Oscillator Section</b>					
Maximum frequency	$C_T = .001$ mfd, $R_T = 2\text{k}\Omega$		300		kHz
Initial accuracy	$R_T$ and $C_T$ constant		5		%
Voltage stability	$V_{IN} = 8$ to $40\text{V}$ , $T_A = 25^\circ\text{C}$			1	%
Temperature stability	Over operating temperature range			2	%
Output amplitude	Pin 3, $T_A = 25^\circ\text{C}$		3.5		$V_P$
Output pulse width	$C_T = .01$ mfd, $T_A = 25^\circ\text{C}$		0.5		$\mu\text{s}$
<b>Error Amplifier Section</b>					
Input offset voltage	$V_{CM} = 2.5\text{V}$		2	10	mV
Input bias current	$V_{CM} = 2.5\text{V}$		2	10	$\mu\text{A}$
Open loop voltage gain		68	80		dB
Common mode voltage	$T_A = 25^\circ\text{C}$	1.8		3.4	V
Common mode rejection ratio	$T_A = 25^\circ\text{C}$		70		dB
Small signal bandwidth	$A_V = 0\text{dB}$ , $T_A = 25^\circ\text{C}$		3		MHz
Output voltage	$T_A = 25^\circ\text{C}$	0.5		3.8	V
<b>Comparator Section</b>					
Duty cycle	% each output "ON"	0		45	%
Input threshold	Zero duty cycle		1		V
Input threshold	Maximum duty cycle		3.5		V
Input bias current			1		$\mu\text{A}$
<b>Current Limiting Section</b>					
Sense voltage	Pin 9 = $2\text{V}$ with error amplifier set for maximum out, $T_A = 25^\circ\text{C}$	180	200	220	mV
Sense voltage T.C.			0.2		$\text{mV}/^\circ\text{C}$
Common mode voltage		-1		+1	V
<b>Output Section (each output)</b>					
Collector-emitter voltage (breakdown)		40			V
Collector-leakage current	$V_{CE} = 40\text{V}$		0.1	50	$\mu\text{A}$
Saturation voltage	$I_C = 50\text{mA}$		1	2	V
Emitter output voltage	$V_{IN} = 20\text{V}$	17	18		V
Rise time	$R_C = 2\text{k}\Omega$ , $T_A = 25^\circ\text{C}$		0.2		$\mu\text{s}$
Fall time	$R_C = 2\text{k}\Omega$ , $T_A = 25^\circ\text{C}$		0.1		$\mu\text{s}$
<b>Total standby current</b> (excluding oscillator charging current, error and current limit dividers, and with outputs open)	$V_{IN} = 40\text{V}$		8	10	mA

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**FM RADIO CIRCUIT**

**TDA 7000**

**Preliminary**

**DESCRIPTION**

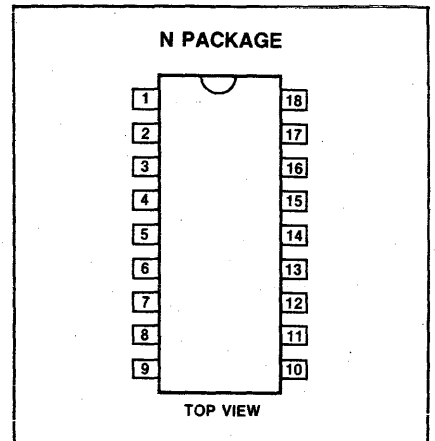
The TDA7000 is a monolithic integrated circuit for mono FM portable radios where a minimum of peripheral components is important (small dimensions and low costs).

The IC has an FLL (Frequency-Locked-Loop) system with an intermediate frequency of 70 kHz. The I.F. selectivity is obtained by active RC filters. The only function which needs alignment is the resonant circuit for the oscillator, thus selecting the reception frequency. Spurious reception is avoided by means of a mute circuit, which also eliminates too-noisy input signals. Special precautions are taken to meet the radiation requirements.

**FEATURES**

- R.F. input stage
- Mixer
- Local oscillator
- I.F. amplifier/limiter
- Phase demodulator
- Mute detector
- Mute switch

**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

SYMBOL AND PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage (pin 5)	12	V
V <sub>6-5</sub> Oscillator voltage (pin 6)	V <sub>CC</sub> -0.5 to V <sub>CC</sub> +0.5	V
Total power dissipation	See derating curve Figure 2	
T <sub>STG</sub> Storage temperature range	-55 to +150	°C
T <sub>A</sub> Operating ambient temperature range	0 to +60	°C

**FUNCTIONAL PIN DESCRIPTION**

PIN NO.	NAME AND FUNCTION
1	Muting capacitor
2	Audio frequency output
3	Noise source
4	Loud filter capacitor
5	Supply voltage
6	VCO
7	1st integrator capacitor (to pin 9)
8	2nd integrator capacitor
9	1st integrator capacitor (to pin 7)
10	IF filter capacitor
11	IF filter capacitor
12	IF limiter capacitor
13	RF input
14	Mixer
15	Current source capacitor
16	Ground
17	Demodulator capacitor
18	Correlator capacitor

LINEAR

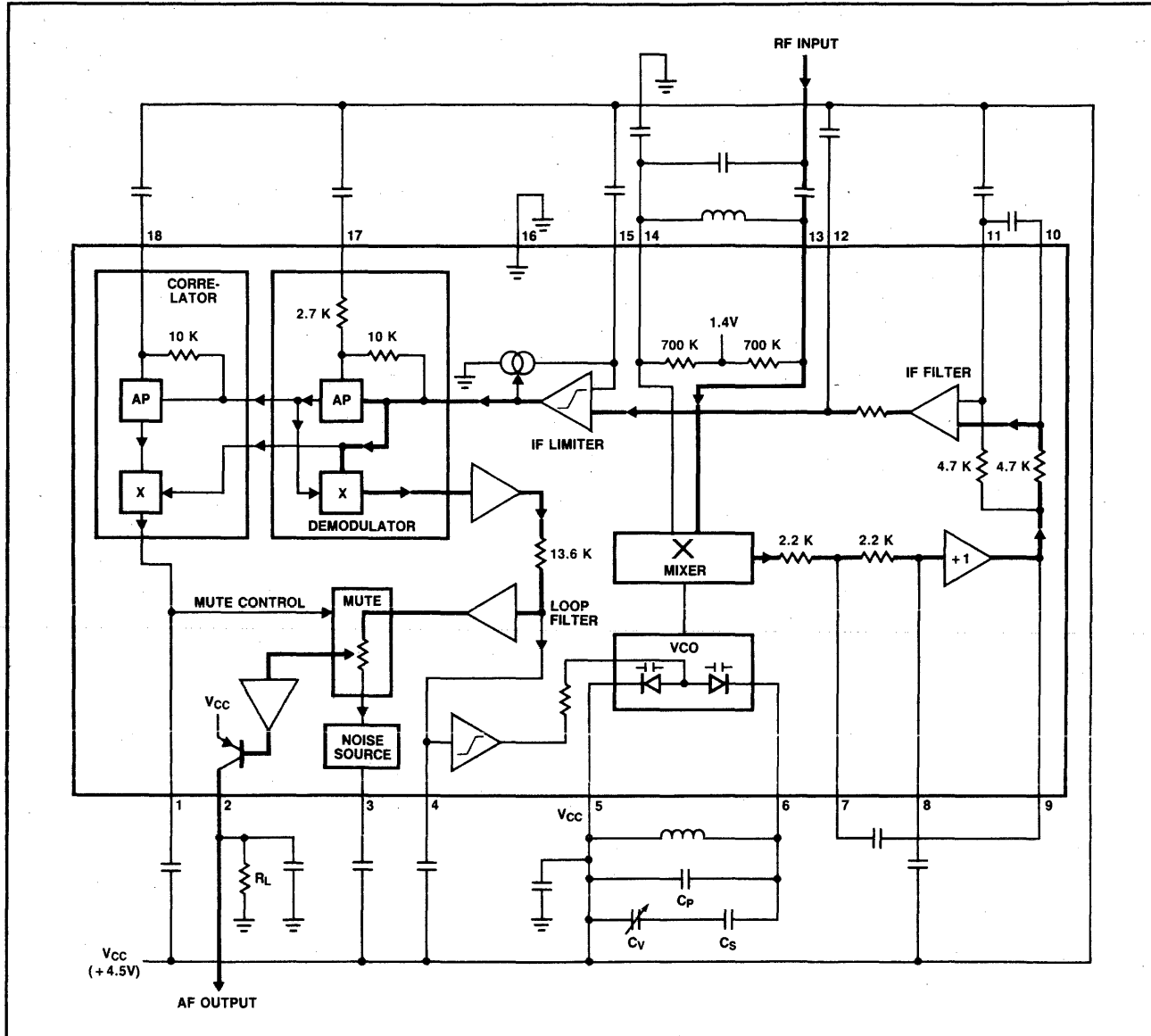
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# FM RADIO CIRCUIT

## TDA 7000

Preliminary

### BLOCK DIAGRAM



LINEAR  
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# FM RADIO CIRCUIT

TDA 7000

**Preliminary**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 4.5V$ ;  $T_A = 25^\circ C$ : measured in Figure 3; unless otherwise specified

SYMBOL AND PARAMETER	TEST CONDITION	TDA7000			UNIT
		Min	Typ	Max	
$V_{CC}$ Supply voltage	(Pin 5)	2.7	4.5	10	V
$I_{CC}$ Supply current	$V_{CC} = 4.5V$		8		mA
$I_6$ Oscillator current	(Pin 6)		280		$\mu A$
$V_{14-16}$ Voltage	(Pin 14)		1.35		V
$I_2$ Output current	(Pin 2)		60		$\mu A$
$V_{2-16}$ Output voltage	(Pin 2) $R_L = 22\ k\Omega$		1.3		V

**AC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 4.5\ V$ ;  $T_A = 25^\circ C$ ; measured in Figure 3 (mute switch open, enabled);  $f_{rf} = 96\ MHz$  (tuned to max. signal at  $5\ \mu V$  e.m.f.) modulated with  $\Delta f = \pm 22.5\ kHz$ ;  $f_m = 1\ kHz$ ;  $EMF = 0.2\ mV$  (e.m.f. voltage at a source impedance of  $75\ \Omega$ ); r.m.s. noise voltage measured unweighted ( $f = 300\ Hz$  to  $20\ kHz$ ); unless otherwise specified.

SYMBOL AND PARAMETER	TEST CONDITION	TDA7000			UNIT
		Min	Typ	Max	
EMF Sensitivity (see Figure 2) (e.m.f. voltage)	-3 dB limiting; muting disabled		1.5		$\mu V$
	-3 dB muting		6		
	S/N = 26 dB		5.5		
EMF Signal handling (e.m.f. voltage)	THD < 10%; $\Delta f = \pm 75\ kHz$		200		mV
S/N Signal-to-noise ratio			60		dB
THD Total harmonic distortion	$\Delta f = \pm 22.5\ kHz$		0.7		%
	$\Delta f = \pm 75\ kHz$		2.3		
AMS AM suppression of output voltage	(ratio of the AM output signal referred to the FM output signal) FM signal: $f_m = 1\ kHz$ ; $\Delta f = \pm 75\ kHz$ AM signal: $f_m = 1\ kHz$ ; $m = 80\%$		50		dB
RR Ripple rejection	( $\Delta V_{CC} = 100\ mV$ ; $f = 1\ kHz$ )		10		dB
$V_{6-5(rms)}$ Oscillator voltage (r.m.s. value)	(Pin 6)		250		mV
$\Delta f_{osc}$ Variation of oscillator frequency	Supply voltage ( $\Delta V_{CC} = 1V$ )		60		kHz/V
$S_{+300}$ Selectivity		45		dB	
		35		dB	
$S_{-300}$					
$\Delta f_{rf}$ A.F.C. range			$\pm 300$		kHz
B Audio bandwidth	$\Delta V_O = 3\ dB$ measured with pre-emphasis ( $t = 50\ \mu s$ )		10		kHz
$V_{O(rms)}$ A.F. output voltage (r.m.s. value)	$R_L = 22\ k\Omega$		75		mV
$R_L$ Load resistance	$V_{CC} = 4.5V$			22	k $\Omega$
	$V_{CC} = 9.0V$			47	

**NOTES:**

- The muting system can be disabled by feeding a current of about  $20\ \mu A$  into pin 1.
- The interstation noise level can be decreased by choosing a low-value capacitor at pin 3. Silent tuning can be achieved by omitting this capacitor.

LINEAR  
Signetics

**FM RADIO CIRCUIT**

**TDA7010T**

**Preliminary**

**DESCRIPTION**

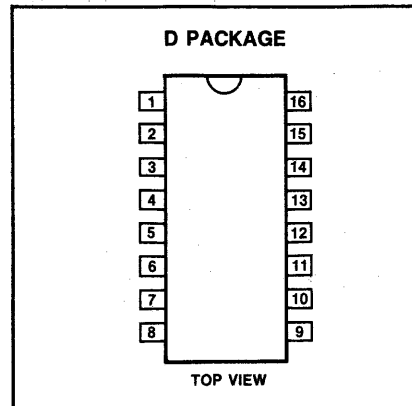
The TDA7010T is a monolithic integrated circuit for mono FM portable radios, where a minimum of peripheral components is important (small dimensions and low costs).

The IC has an FLL (Frequency-Locked-Loop) system with an intermediate frequency of 70 kHz. The I.F. selectivity is obtained by active RC filters. The only function which needs alignment is the resonant circuit for the oscillator, thus selecting the reception frequency. Spurious reception is avoided by means of a mute circuit, which also eliminates too noisy input signals. Special precautions are taken to meet the radiation requirements.

**FEATURES**

- R.F. input stage
- Mixer
- Local oscillator
- I.F. amplifier/limiter
- Phase demodulator
- Mute detector
- Mute switch

**PIN CONFIGURATION**



**FUNCTIONAL PIN DESCRIPTION**

PIN NO.	NAME AND FUNCTION
1	Muting capacitor
2	Audio frequency output
3	Loop filter capacitor
4	Supply voltage
5	VCO
6	1st integrator capacitor (to pin 8)
7	2nd integrator capacitor
8	1st integrator capacitor (to pin 6)
9	IF filter capacitor
10	IF limiter capacitor
11	RF input
12	Mixer
13	Current source capacitor
14	Ground
15	Demodulator capacitor
16	Correlator capacitor

**ABSOLUTE MAXIMUM RATINGS**

SYMBOL AND PARAMETER	RATING	UNIT
$V_{CC}$ Supply voltage (pin 4)	12	V
$V_{6-5}$ Oscillator voltage (pin 5)	$V_{CC}-0.5$ to $V_{CC}+0.5$	V
Total power dissipation	See derating curve Figure 2	
$T_{STG}$ Storage temperature range	-55 to +150	°C
$T_A$ Operating ambient temperature range	0 to +60	°C

LINEAR

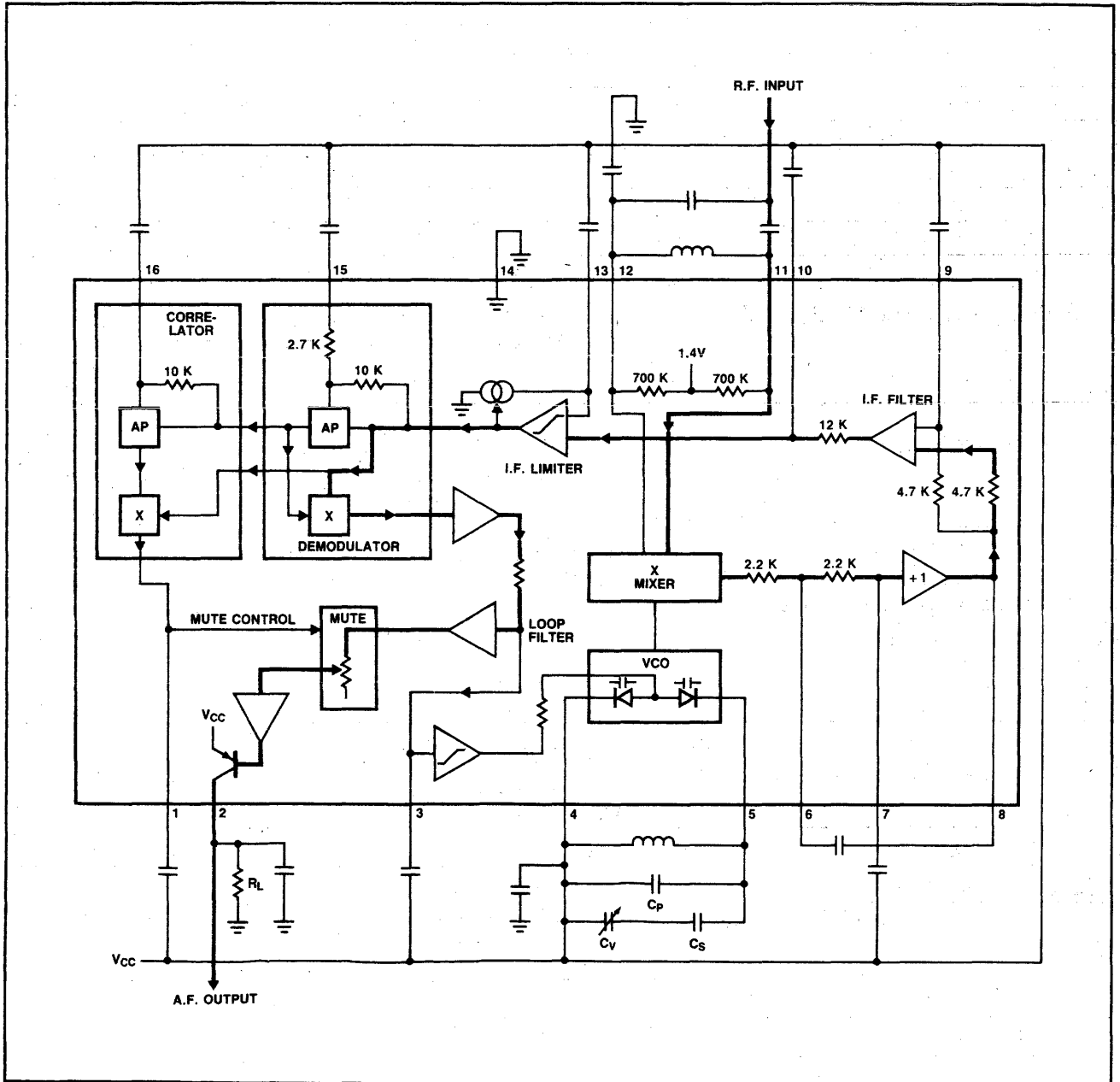
Signetics

# FM RADIO CIRCUIT

TDA7010T

Preliminary

## BLOCK DIAGRAM



LINEAR

Signetics

## FM RADIO CIRCUIT

TDA7010T

## Preliminary

DC ELECTRICAL CHARACTERISTICS  $V_{CC} = 4.5V$ ;  $T_A = 25^\circ C$ ; measured in Figure 3; unless otherwise specified.

SYMBOL AND PARAMETER	TEST CONDITION	TDA3810			UNIT
		Min	Typ	Max	
$V_{CC}$ Supply voltage	(Pin 4)	2.7	4.5	10	V
$I_{CC}$ Supply current	$V_{CC} = 4.5V$		8		mA
$I_5$ Oscillator Current	(Pin 5)		280		$\mu A$
$V_{12-14}$ Voltage	(Pin 12)		1.35		V
$I_2$ Output current	(Pin 2)		60		$\mu A$
$V_{2-14}$ Output voltage	(Pin 2) $R_L = 22 k\Omega$		1.3		V

AC ELECTRICAL CHARACTERISTICS  $V_{CC} = 4.5 V$ ;  $T_A = 25^\circ C$ ; measured in Figure 3 (mute switch open, enabled);  $f_r = 96$  MHz (tuned to max. signal at  $5 \mu V$  e.m.f.) modulated with  $\Delta f = \pm 22.5$  kHz;  $f_m = 1$  kHz; EMF = 0.2 mV (e.m.f. voltage at a source impedance of  $75 \Omega$ ); r.m.s. noise voltage measured unweighted ( $f = 300$  Hz to 20 kHz); unless otherwise specified.

SYMBOL AND PARAMETER	TEST CONDITION	TDA7010T			UNIT
		Min	Typ	Max	
EMF Sensitivity (see Figure 2) (e.m.f. voltage)	-3 dB limiting; muting disabled		1.5		$\mu V$
	-3 dB muting		6		
	S/N = 26 dB		5.5		
EMF Signal handling (e.m.f. voltage)	THD < 10%; $\Delta f = \pm 75$ kHz		200		mV
S/N Signal-to-noise ratio			60		dB
THD Total harmonic distortion	$\Delta f = \pm 22.5$ kHz $\Delta f = \pm 75$ kHz		0.7		%
			2.3		%
AMS AM suppression of output voltage	(ratio of the AM output signal referred to the FM output signal) FM signal: $f_m = 1$ kHz; $\Delta f = \pm 75$ kHz AM signal: $f_m = 1$ kHz; $m = 80\%$		50		dB
RR Ripple rejection	( $\Delta V_{CC} = 100$ mV; $f = 1$ kHz)		10		dB
$V_{5-4(rms)}$ Oscillator voltage (r.m.s. value)	(Pin 5)		250		mV
$\Delta f_{osc}$ Variation of oscillator frequency	Supply voltage ( $\Delta V_{CC} = 1V$ )		60		kHz/V
$S_{+300}$ $S_{-300}$ Selectivity		43			dB
		28			
$\Delta f_{rf}$ A.F.C. range			$\pm 300$		kHz
B Audio bandwidth	$\Delta V_O = 3$ dB measured with pre-emphasis ( $t = 50 \mu s$ )		10		kHz
$V_{O(rms)}$ A.F. output voltage (r.m.s. value)	$R_L = 22 k\Omega$		75		mV
$R_L$ Load resistance	$V_{CC} = 4.5V$			22	k $\Omega$
	$V_{CC} = 9.0V$			47	

**SAMPLE AND HOLD AMPLIFIER**

**NE/SE5537**

**DESCRIPTION**

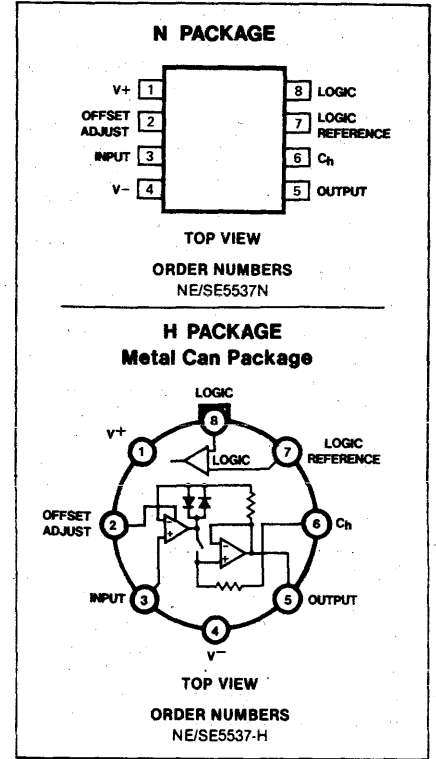
The NE5537 monolithic Sample and Hold amplifier combines the best features of ion implanted JFET's with bipolar devices to obtain high accuracy, fast acquisition time, and low droop rate. This device is pin compatible with the LF198, and features superior performance in droop rate and output drive capability. The circuit shown in Figure 1 contains two operational amplifiers which function as a unity gain amplifier in the Sample mode. The first amplifier has bipolar input transistors which gives the system a low offset voltage. The second amplifier has JFET input transistors to achieve low leakage current from the hold capacitor. A unique circuit design for leakage current cancellation using current mirrors gives the NE5537 a low droop rate at higher temperature. The output stage has the capability to drive a 2KΩ load. The logic input is compati-

ble with TTL, PMOS or CMOS logic. The differential logic threshold is 1.4V with the Sample mode occurring when the logic input is high. It is available in 8-lead TO-5 and 8-pin plastic DIP packages.

**FEATURES**

- Operates from ±5V to ±18V supplies
- Hold leakage current 6pA @ T<sub>J</sub>25°C
- Less than 10μs acquisition time
- TTL, PMOS, CMOS compatible logic input
- 0.5mV typical hold step at C<sub>H</sub> = 0.01μF
- Low input offset: 1mV (typical)
- 0.002% gain accuracy with R<sub>L</sub> = 2kΩ
- Low output noise in hold mode
- Input characteristics do not change during hold mode
- High supply rejection ratio in sample or hold
- Wide bandwidth

**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
Supply voltage	± 18	V
Power dissipation (package limitation) <sup>1</sup>	500	mW
Operating ambient temperature range		
SE5537	-55 to +125	°C
NE5537	0 to +70	°C
Storage temperature range	-65 to +150	°C
Input voltage	Equal to supply voltage	
Logic to logic reference differential voltage <sup>2</sup>	+7, -30	V
Output short circuit duration	Indefinite	
Hold capacitor short circuit duration	10	sec
Lead temperature (soldering, 10sec)	300	°C

**NOTES**

1. The maximum junction temperature of the SE5537 is 150°C and for the NE5537 is 100°C. When operating at elevated ambient temperature, the TO-5 and plastic DIP packages must be derated based on a thermal resistance (θ<sub>JA</sub>) of 150°C/W.
2. Although the differential voltage may not exceed the limits given, the common mode voltage on the logic pins may be equal to the supply voltages without causing damage to the circuit. For proper logic operation, however, one of the logic pins must always be at least 2V below the positive supply and 3V above the negative supply.

**BLOCK DIAGRAM**

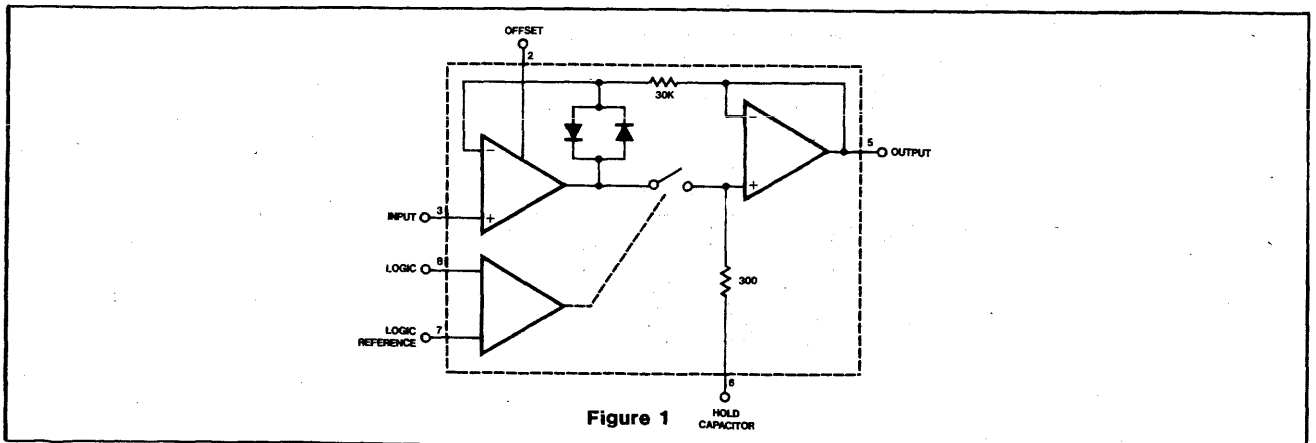


Figure 1

QUAD TIMER

NE/SA/SE558

DESCRIPTION

The 558 Quad Timers are monolithic timing devices which can be used to produce four entirely independent timing functions. The 558 output sinks current. These highly stable, general purpose controllers can be used in a monostable mode to produce accurate time delays, from microseconds to hours. In the time delay mode of operation, the time is precisely controlled by one external resistor and one capacitor. A stable operation can be achieved by using two of the four timer sections.

The four timing sections in the 558 are edge triggered; therefore, when connected in tandem for sequential timing applications, no coupling capacitors are required. Output current capability of 100mA is provided in both devices.

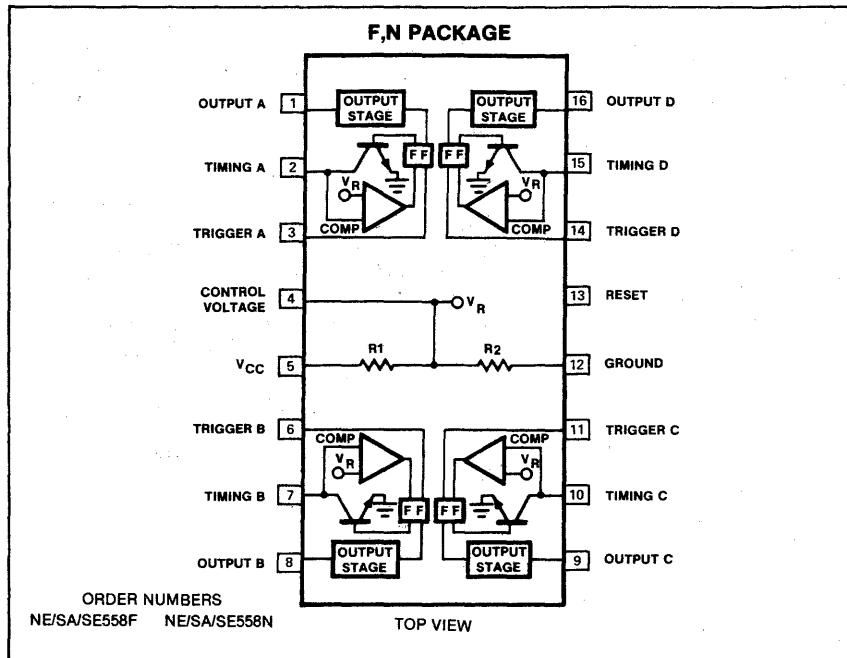
FEATURES

- 100mA output current per section
- Edge triggered (no coupling capacitor)
- Output independent of trigger conditions
- Wide supply voltage range 4.5V to 18V
- Timer intervals from microseconds to hours
- Time period equals RC
- Military qualifications pending

APPLICATIONS

- Sequential timing
- Time delay generation
- Precision timing
- Industrial controls
- Quad one-shot

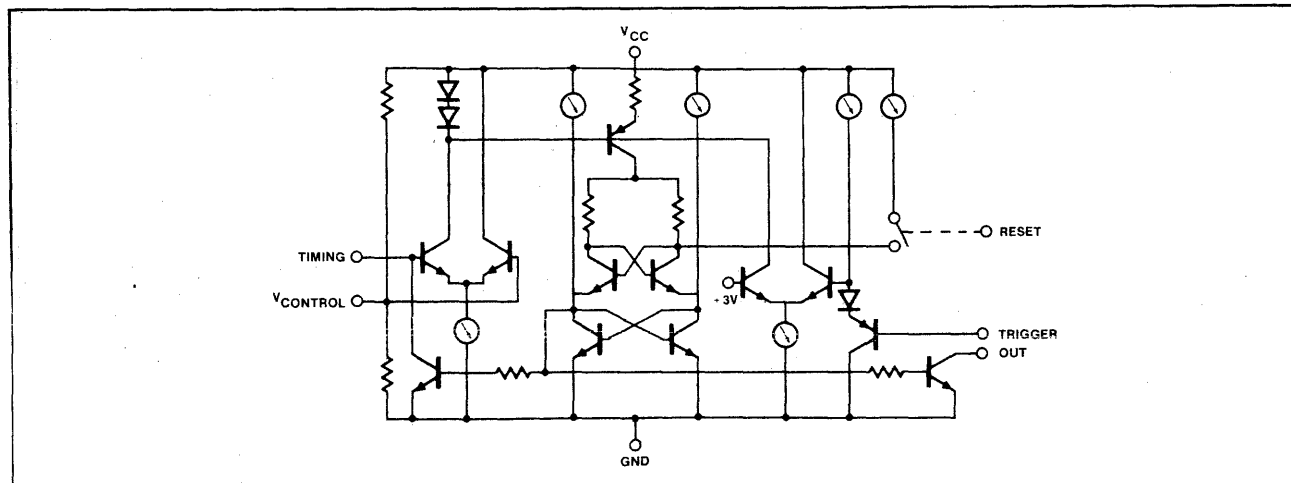
PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Supply voltage		
NE/SA558	+16	V
SE558	+18	V
Power dissipation	1.25	W
Operating temperature range		
NE558	0 to +70	°C
SA558	-40 to +85	°C
SE558	-55 to +125	°C
Storage temperature range	-65 to +150	°C
Lead temperature (soldering, 60sec)	+300	°C

558 EQUIVALENT CIRCUIT





## QUAD TIMER

NE/SA/SE558

ELECTRICAL CHARACTERISTICS  $T_A = 25^\circ\text{C}$ ,  $V_{CC} = +5\text{V}$  to  $+15\text{V}$  unless otherwise specified.

PARAMETER	TEST CONDITIONS	SE558			NE/SA558			UNIT
		Min	Typ	Max	Min	Typ	Max	
Supply voltage		4.5		18	4.5		16	V
Supply current	$V_{CC} = \text{Reset} = 15\text{V}$		16	32		16	36	mA
Timing accuracy ( $T = RC$ )	$R = 2\text{k}\Omega$ to $100\text{k}\Omega$ $C = 1\mu\text{F}$							
Initial accuracy			$\pm 1.0$	3		$\pm 2$	5	%
Drift with temperature			30	100		30	150	ppm/ $^\circ\text{C}$
Drift with supply voltage			0.1	0.9		0.1	0.9	%/V
Trigger voltage <sup>1</sup>	$V_{CC} = 15\text{V}$	0.8		2.4	0.8		2.4	V
Trigger current	Trigger = 0V		5	30		5	100	$\mu\text{A}$
Reset voltage <sup>2</sup>		0.8		2.4	0.8		2.4	V
Reset current	Reset		50	300		50	500	$\mu\text{A}$
Threshold voltage			0.63			0.63		$\times V_{CC}$
Threshold leakage			15			15		nA
Output voltage <sup>3</sup>	$I_L = 10\text{mA}$		0.1	0.2		0.1	0.4	V
	$I_L = 100\text{mA}$		0.7	1.5		1.0	2.0	V
Output leakage			10	500		10	500	nA
Propagation delay			1.0			1.0		$\mu\text{s}$
Risetime of output	$I_L = 100\text{mA}$		100			100		ns
Falltime of output	$I_L = 100\text{mA}$		100			100		ns

## NOTES

1. The trigger functions only on the falling edge of the trigger pulse only after previously being high. After reset the trigger must be brought high and then low to implement triggering.
2. For reset below 0.8 volts, outputs set low and trigger inhibited. For reset above 2.4 volts, trigger enabled.
3. The 558 output structure is open collector which requires a pull up resistor to  $V_{CC}$  to sink current. The output is normally low sinking current.

VOICE SYNTHESIZER

MEA8000

Preliminary

DESCRIPTION

The MEA8000 is a 24-pin N MOS integrated circuit for generating good quality speech from digital code with a programmable bit rate. The circuit is primarily intended for applications in microprocessor controlled systems, where the speech code is stored separately in a Read-Only Memory. An efficient, easy-to-use speech editing and encoding system with EPROM programming capability, has been specially developed.

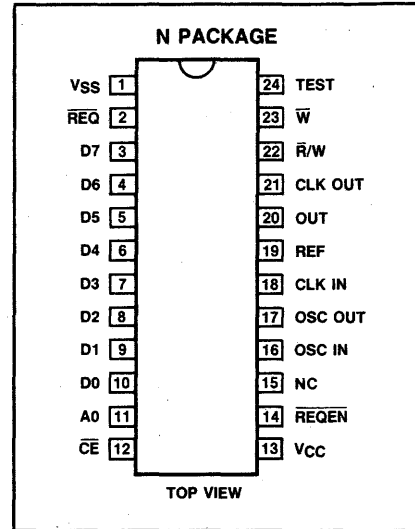
APPLICATIONS

- Telephony
- Automotive
- Computer response/prompt.
- Video games.
- General industrial.

FEATURES

- Microprocessor interface capability including an 8-bit data bus, an enable and a read/write input control signals.
- 32-bit data buffer holding speech frame codes.
- Digital filter of 8th order with 3 programmable formant frequencies, one fixed formant frequency, and 4 programmable formant bandwidths. 4 KHZ bandwidth.
- Programmable amplitudes.
- Programmable duration of each frame: 8, 16, 32, or 64 milliseconds.
- Low data rate: average 1000 bits/sec.
- Operates from standard EPROMs/ROMs.
- Minimal external audio filter requirement.
- Crystal controlled oscillator or external (TTL) clock.
- Single +5V power supply.

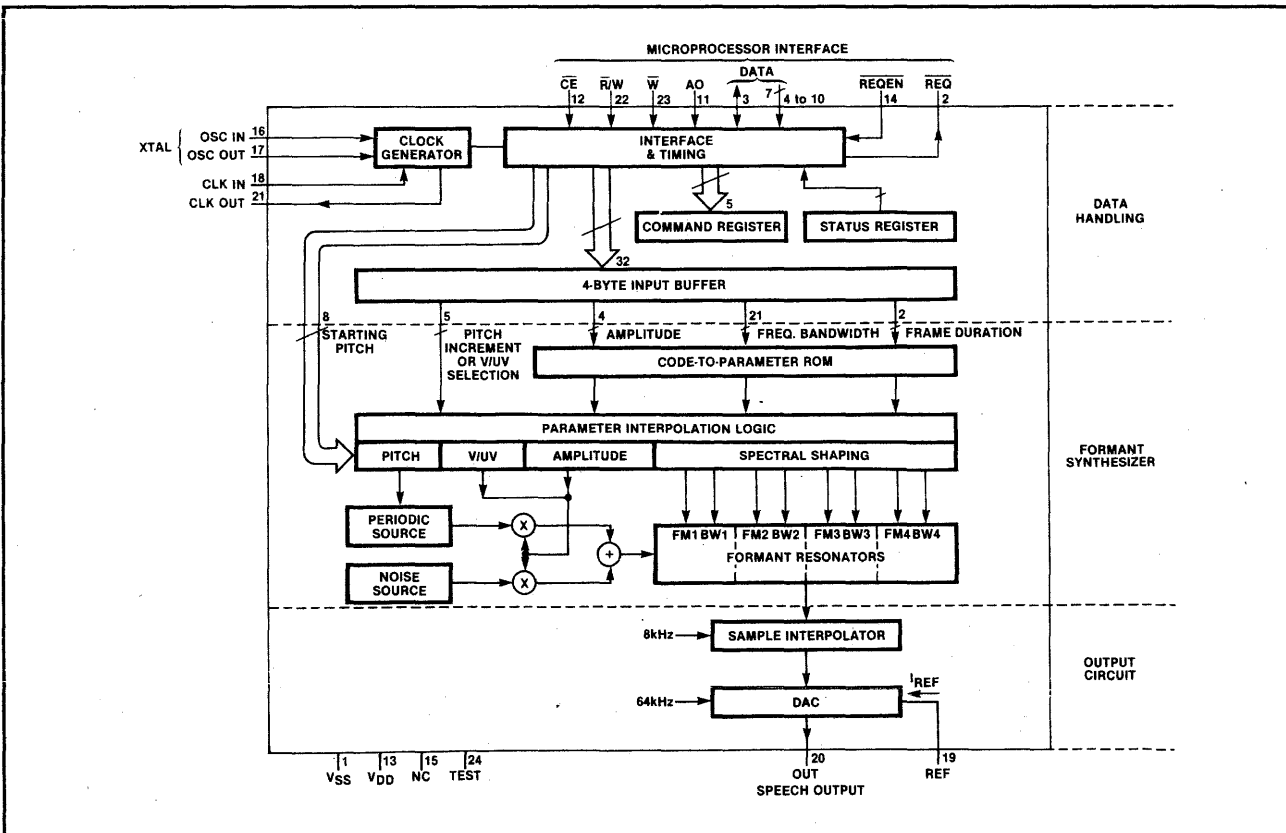
PIN CONFIGURATION



ABSOLUTE MAXIMUM RATING

SYMBOL AND PARAMETER	RATING	UNIT
V <sub>CC</sub>	Supply voltage	-0.5 to +7 V
V <sub>I</sub>	Voltage on any pin with respect to V <sub>SS</sub>	-0.5 to +7 V
V <sub>REQ</sub> , V <sub>OUT</sub>	Output voltage on pins 2 and 20	15 V
T <sub>STG</sub>	Storage temperature range	-20 to +125 °C
T <sub>A</sub>	Operating ambient temperature range	0 to +70 °C

BLOCK DIAGRAM



LINEAR

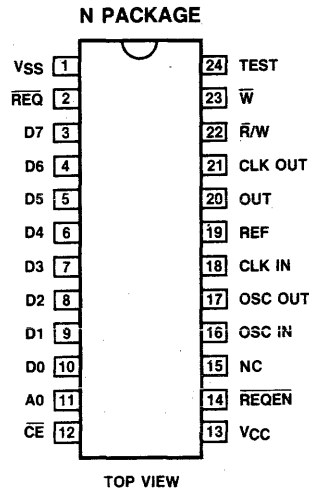
Signetics

**VOICE SYNTHESIZER**

**MEA8000**

**Preliminary**

**FUNCTIONAL PIN DESCRIPTION**



PIN NO.	SYMBOL	NAME AND FUNCTION
<b>CONTROL</b>		
2	$\bar{REQ}$	<b>DATA REQUEST</b> output signal (open drain) which follows the inverse of the status REQ bit, but only if enabled by either the ROE bit in the command register or the external REQEN input pin.
3 to 10	D7 to D0	<b>Data bus</b> to which command or encoded speech parameters can be written. D7 is a bidirectional line through which the status bit can be read.
11	A0	<b>Data/control</b> input. Discriminates between speech code input buffer (A0 = '0') and command register (A0 = '1') during a 'write' operation.
12	$\bar{CE}$	<b>Chip enable, Read/Write, Write</b> These control signals provides an easy interface to most microprocessors or microcomputers (see timing diagrams).
22	$\bar{R}/\bar{W}$	
23	$\bar{W}$	
14	REQEN	<b>Request enable</b> input. REQEN = '0' enables the status REQ bit to appear inverted on the REQ output, independent of the command register.
<b>TIMING</b>		
16	OSC IN	Connections for internal clock oscillator. Nominal crystal frequency is 3.84 MHz. OSC IN must be tied to ground if CLK IN is used.
17	OSC OUT	
18	CLK IN	<b>Clock input</b> for external clock, TTL compatible, 3.84 MHz. Must be tied to ground when not used.
21	CLK OUT	A buffered output of the internal clock cycle ( = CLK IN divided by 3).
<b>OUTPUT</b>		
19	REF	<b>Reference Current</b> input pin for biasing the audio output level. This reference current can be derived from a resistor to the positive supply.
20	OUT	<b>Speech output.</b> This output is a 64 kHz pulse, modulated in both width and amplitude. It is configured as a current sink with a saturating voltage of about 3V.
<b>SUPPLY</b>		
1	V <sub>SS</sub>	<b>Ground.</b>
13	V <sub>CC</sub>	<b>Single supply voltage.</b> Nominally 5V, but battery operation is also possible.
15	NC	No connection.
24	TEST	Used for testing purposes. Changes other pin functions. Must be tied to ground for user operation.

**LINEAR**  
**Signetics**

# VOICE SYNTHESIZER

## MEA8000

### Preliminary

DC ELECTRICAL CHARACTERISTICS  $T_A = 25^\circ\text{C}$ :  $V_{CC} = 5\text{V}$ , unless otherwise stated. All voltages referenced to  $V_{SS}$

SYMBOL AND PARAMETER		TEST CONDITION	MEA8000			UNIT
			Min	Typ	Max	
$V_{CC}$	Supply voltage (note 1)	(No audio load)	4.5	5.0	5.5	V
$I_{CC}$	Supply current		30	50	mA	
$V_{IH}$	D0 to D7, A0, CE, W R/W REGEN, CLK IN: Input HIGH voltage		2.0		$V_{CC}$	V
$V_{IL}$	Input LOW voltage		-0.5		0.8	V
$I_{IR}$	Input Leakage current (note 2)				10	$\mu\text{A}$
$C_I$	Input Capacitance				7	pF
$V_{OH}$	D7 (I/O), CLK OUT: Output HIGH voltage	$-I_{OH} = -100 \mu\text{A}$ $I_{OL} = 1.6 \text{ mA}$	2.4			V
$V_{OL}$	Output LOW voltage				0.4	V
$C_L$	Output Load capacitance				50	pF
	REQ:	Open drain $I_{OL} = 1.6 \text{ mA}$				
$V_{OH}$	Output HIGH voltage				13.2	V
$V_{OL}$	Output LOW voltage				0.4	V
$C_L$	Output Load capacitance			50	pF	
	Audio output					
$I_{REF}$	Reference current (note 8) - Pin 19				0.3	mA
$I_{OUT}$	Output current (peak) - Pin 20					
			$I_{REF} = 0 \text{ mA}$ $I_{REF} = 0.1 \text{ mA}$ $I_{REF} = 0.3 \text{ mA}$ $I_{REF} = 0.1 \text{ mA}$	100 1.7 5		$\mu\text{A}$ mA mA
$V_{OUT}$	$V_{OUT}$ (pin 20) for linear operation (note 3)		2.5		13.2	V
	Oscillator	Internal External				
$f_{XTAL}$	Crystal frequency			3.84	4.00	MHz
$f_{CLK}$	Clock frequency		3.84	4.00	MHz	

### AC ELECTRICAL CHARACTERISTICS (note 4) (Figure 4 and 5)

SYMBOL AND PARAMETER		TEST CONDITION	MEA8000			UNIT
			Min	Typ	Max	
$t_{WR}$	Write enable pulse width.	Clock frequency = 3.84 MHz	200			ns
$t_{AS}$	Address set-up time.		30			ns
$t_{AH}$	Address hold time.		30			ns
$t_{DS}$	Data set-up time for write operation.		150			ns
$t_{DH}$	Data hold time for write operation.		30			ns
$t_{RH}$	Request hold time (note 5)				350	ns
$t_{RN}$	Request next (note 6)				3	$\mu\text{s}$
$t_{RD}$	Read enable time.		200			ns
$t_{DD}$	Data delay for read operation (note 7)				150	ns
$t_{DF}$	Data floating for read operation (note 7)				150	ns
$t_{RV}$	Request valid before a write operation.		0			ns
$t_{ROE}$	Request output enable response.				750	ns
$t_{CS}$	Control set-up time.				20	ns
$t_{CH}$	Control hold time.				20	ns

#### NOTES

- The circuit will continue to operate from a supply of up to 6.5V, but without necessarily meeting the specification.
- This is also valid for  $V_{CC} = 0\text{V}$ .
- This permits connection of the output load to a supply higher than that supplying the synthesizer.
- Timing reference level is 1.5V.
- An external pull up resistor is required, as this is an open drain output. The time ( $t_{RH}$ ) to reach 2.0V is specified at a load to 5V of 3.3 k $\Omega$  and 50 pF.
- Between two data write operations of one speech frame.
- Levels greater than 2.0V for a '1' or less than 0.8V for a '0' are reached with a load of one TTL input and 50 pF.
- Typical voltage level at the REF pin is 2.5V.

**VIDEO AMPLIFIER**

**NE/SE592**

**DESCRIPTION**

The SE/NE592 is a monolithic, two stage, differential output, wideband video amplifier. It offers fixed gains of 100 and 400 without external components and adjustable gains from 400 to 0 with one external resistor. The input stage has been designed so that with the addition of a few external reactive elements between the gain select terminals, the circuit can function as a high pass, low pass, or band pass filter. This feature makes the circuit ideal for use as a video or pulse amplifier in communications, magnetic memories, display, video recorder systems, and floppy disk head amplifiers. Now available in an 8-pin version with fixed gain of 400 without external components and adjustable gain from 400 to 0 with one external resistor.

**FEATURES**

- 120MHz bandwidth
- Adjustable gains from 0 to 400
- Adjustable pass band
- No frequency compensation required
- Wave shaping with minimal external components

**APPLICATIONS**

- Floppy disk head amplifier
- Video amplifier
- Pulse amplifier in communications
- Magnetic memory
- Video recorder systems

**ABSOLUTE MAXIMUM RATINGS**  $T_A = 25^\circ\text{C}$  unless otherwise specified.

PARAMETER	RATING	UNIT
Supply voltage	$\pm 8$	V
Differential input voltage	$\pm 5$	V
Common mode		
Input voltage	$\pm 6$	V
Output current	10	mA
Operating temperature range		
SE592	-55 to +125	$^\circ\text{C}$
NE592	0 to +70	$^\circ\text{C}$
Storage temperature range	-65 to +150	$^\circ\text{C}$
Power dissipation	500	mW

**PIN CONFIGURATION**

**DH, FH, N14 PACKAGE**

**TOP VIEW**

**ORDER NUMBERS**  
 NE592DH NE592N14  
 NE592FH SE592FH

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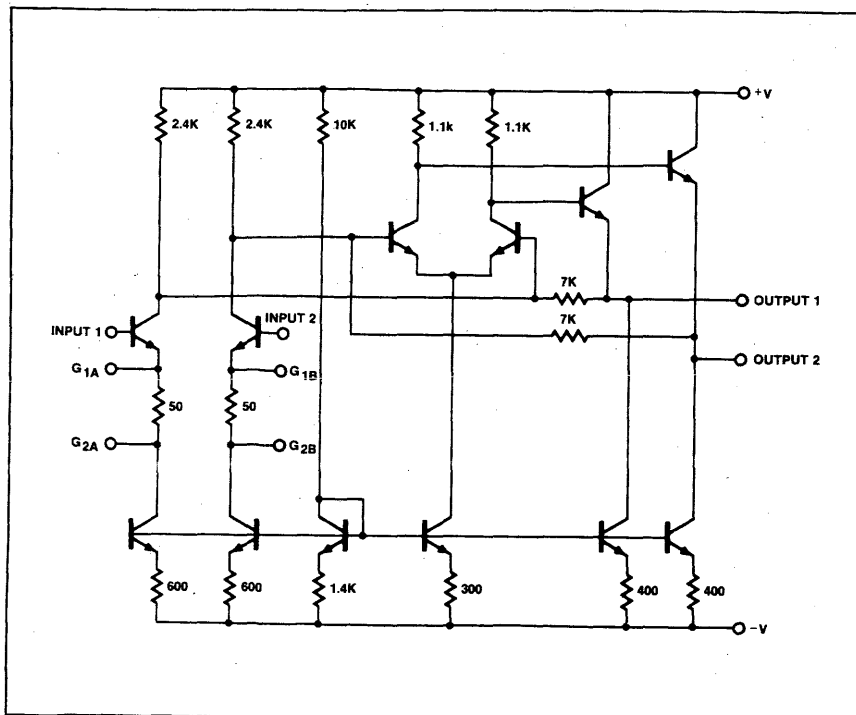
**H PACKAGE\***

**NOTE**  
Pin 5 connected to case.

**ORDER NUMBERS**  
 NE592H SE592H

\*Metal cans (H) not recommended for new designs.

**EQUIVALENT CIRCUIT**



**DE, FE, N8 PACKAGE**

**TOP VIEW**

**ORDER NUMBERS**  
 NE592DE NE592N8  
 NE592FE S592FE

**VIDEO AMPLIFIER**

**NE/SE592**

**DC ELECTRICAL CHARACTERISTICS**  $T_A = +25^\circ\text{C}$ ,  $V_{SS} = \pm 6\text{V}$ ,  $V_{CM} = 0$  unless otherwise specified.  
Recommended operating supply voltages  $V_S = \pm 6.0\text{V}$ .

PARAMETER	TEST CONDITIONS	NE592			SE592			UNITS
		Min	Typ	Max	Min	Typ	Max	
Differential voltage gain Gain 1 <sup>1</sup> Gain 2 <sup>2,4</sup>	$R_L = 2\text{k}\Omega$ , $V_{OUT} = 3\text{V p-p}$	250 80	400 100	600 120	300 90	400 100	500 110	V/V V/V
Bandwidth Gain 1 <sup>1</sup> Gain 2 <sup>2,4</sup> Rise time Gain 1 <sup>1</sup> Gain 2 <sup>2,4</sup>	$V_{OUT} = 1\text{V p-p}$		40 90			40 90		MHz MHz ns ns
Propagation delay Gain 1 <sup>1</sup> Gain 2 <sup>2,4</sup>	$V_{OUT} = 1\text{V p-p}$		7.5 6.0	10		7.5 6.0	10	ns ns
Input resistance Gain 1 <sup>1</sup> Gain 2 <sup>2,4</sup> Input capacitance <sup>2</sup> Input offset current Input bias current Input noise voltage Input voltage range	Gain 2 <sup>4</sup>  BW 1kHz to 10MHz	10	4.0 30 2.0 0.4 9.0 12	5.0 30  $\pm 1.0$		4.0 30 2.0 0.4 9.0 12	3.0 20  $\pm 1.0$	k $\Omega$ k $\Omega$ pF $\mu\text{A}$ $\mu\text{A}$ $\mu\text{Vrms}$ V
Common mode rejection ratio Gain 2 <sup>4</sup> Gain 2 <sup>4</sup> Supply voltage rejection ratio Gain 2 <sup>4</sup>	$V_{CM} \pm 1\text{V}$ , $F < 100\text{kHz}$ $V_{CM} \pm 1\text{V}$ , $F = 5\text{MHz}$  $\Delta V_S = \pm 0.5\text{V}$	60  50	86 60 70		60  50	86 60 70		dB dB dB
Output offset voltage Gain 1 Gain 2 <sup>4</sup> Gain 3 <sup>3</sup> Output common mode voltage Output voltage swing differential Output resistance Power supply current	$R_L = \infty$ $R_L = \infty$ $R_L = \infty$ $R_L = \infty$ $R_L = 2\text{K}$  $R_L = \infty$		0.35 2.9 4.0 20 18	1.5 1.5 0.75 3.4  24		0.35 2.9 4.0 20 18	1.5 1.0 0.75 3.4  24	V V V V V $\Omega$ mA
<b>THE FOLLOWING SPECS APPLY OVER TEMPERATURE</b>		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$			$-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$			
Differential voltage gain Gain 1 <sup>1</sup> Gain 2 <sup>2,4</sup>	$R_L = 2\text{k}\Omega$ , $V_{OUT} = 3\text{V p-p}$	250 80		600 120	200 80		600 120	V/V V/V
Input resistance Gain 2 <sup>2,4</sup> Input offset current Input bias current Input voltage range		8.0  $\pm 1.0$		6.0 40	8.0  $\pm 1.0$		5.0 40	k $\Omega$ $\mu\text{A}$ $\mu\text{A}$ V
Common mode rejection ratio Gain 2 <sup>4</sup> Supply voltage rejection ratio Gain 2 <sup>4</sup>	$V_{CM} \pm 1\text{V}$ , $F < 100\text{kHz}$  $\Delta V_S = \pm 0.5\text{V}$	50  50			50  50			dB dB
Output offset voltage Gain 1 Gain 2 <sup>4</sup> Gain 3 <sup>3</sup> Output voltage swing differential Power supply current	$R_L = \infty$ $R_L = \infty$ $R_L = \infty$ $R_L = 2\text{K}$ $R_L = \infty$	2.8		1.5 1.5 1.0 27	2.5		1.5 1.2 1.0 27	V V V V mA

**NOTES:**

- Gain select pins G<sub>1A</sub> and G<sub>1B</sub> connected together.
- Gain select pins G<sub>2A</sub> and G<sub>2B</sub> connected together.
- All gain select pins open.
- Applies to 14-pin version only.

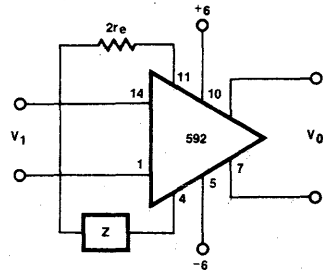
Signetics LINEAIR

**VIDEO AMPLIFIER**

**NE/SE592**

**TYPICAL APPLICATIONS**

**FILTER NETWORKS**



$$\frac{V_0(s)}{V_1(s)} \approx \frac{1.4 \times 10^4}{Z(s) + 2r_e}$$

$$\approx \frac{1.4 \times 10^4}{Z(s) + 32}$$

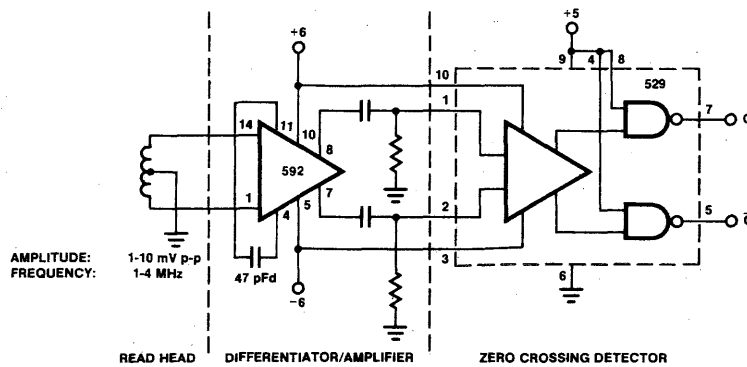
**BASIC CONFIGURATION**

Z NETWORK	FILTER TYPE	$V_0(s)$ TRANSFER $V_1(s)$ FUNCTION
	LOW PASS	$\frac{1.4 \times 10^4}{L} \left[ \frac{1}{s + R/L} \right]$
	HIGH PASS	$\frac{1.4 \times 10^4}{R} \left[ \frac{s}{s + 1/RC} \right]$
	BAND PASS	$\frac{1.4 \times 10^4}{L} \left[ \frac{s}{s^2 + R/L s + 1/LC} \right]$
	BAND REJECT	$\frac{1.4 \times 10^4}{R} \left[ \frac{s^2 + 1/LC}{s^2 + 1/LC + s/RC} \right]$

**NOTE**

In the networks above, the R value used is assumed to include  $2r_e$ , or approximately  $32\Omega$ .

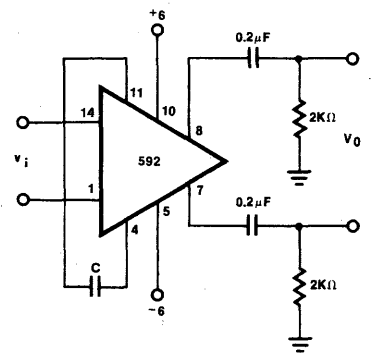
**DISC/TAPE PHASE MODULATED READBACK SYSTEMS**



AMPLITUDE: 1-10 mV p-p  
FREQUENCY: 1-4 MHz

**READ HEAD      DIFFERENTIATOR/AMPLIFIER      ZERO CROSSING DETECTOR**

**DIFFERENTIATION WITH HIGH COMMON MODE NOISE REJECTION**



FOR FREQUENCY  $F_1 \ll 1/2 \pi (32) C$   
 $V_0 \approx 1.4 \times 10^4 C \frac{dV_i}{dT}$

**ULTRA HIGH FREQUENCY OPERATIONAL AMPLIFIER**

**NE/SE5539**

**DESCRIPTION**

The Signetics NE5539 is a very wide bandwidth, high slew rate, monolithic operational amplifier for use in video amplifiers, RF amplifiers, and extremely high slew rate amplifiers.

Emitter follower inputs provide a true differential high input impedance device. Proper external compensation will allow design operation over a wide range of closed loop gains, both inverting and non-inverting, to meet specific design requirements.

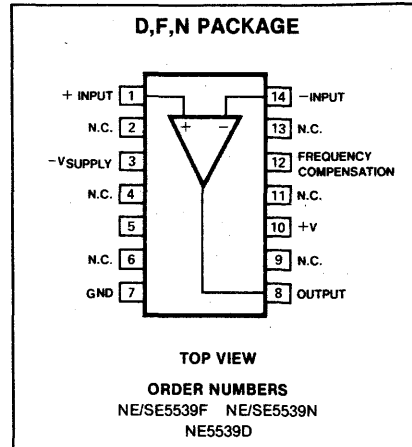
**FEATURES**

- Gain bandwidth product: 1.2GHz at 17dB
- Slew rate: 600V/ $\mu$ sec
- Full power response: 48MHz
- AvOL: 52dB typical

**APPLICATIONS**

- Fast pulse amplifiers
- RF oscillators
- Fast sample and hold
- High gain video amplifiers (BW > 20MHz)

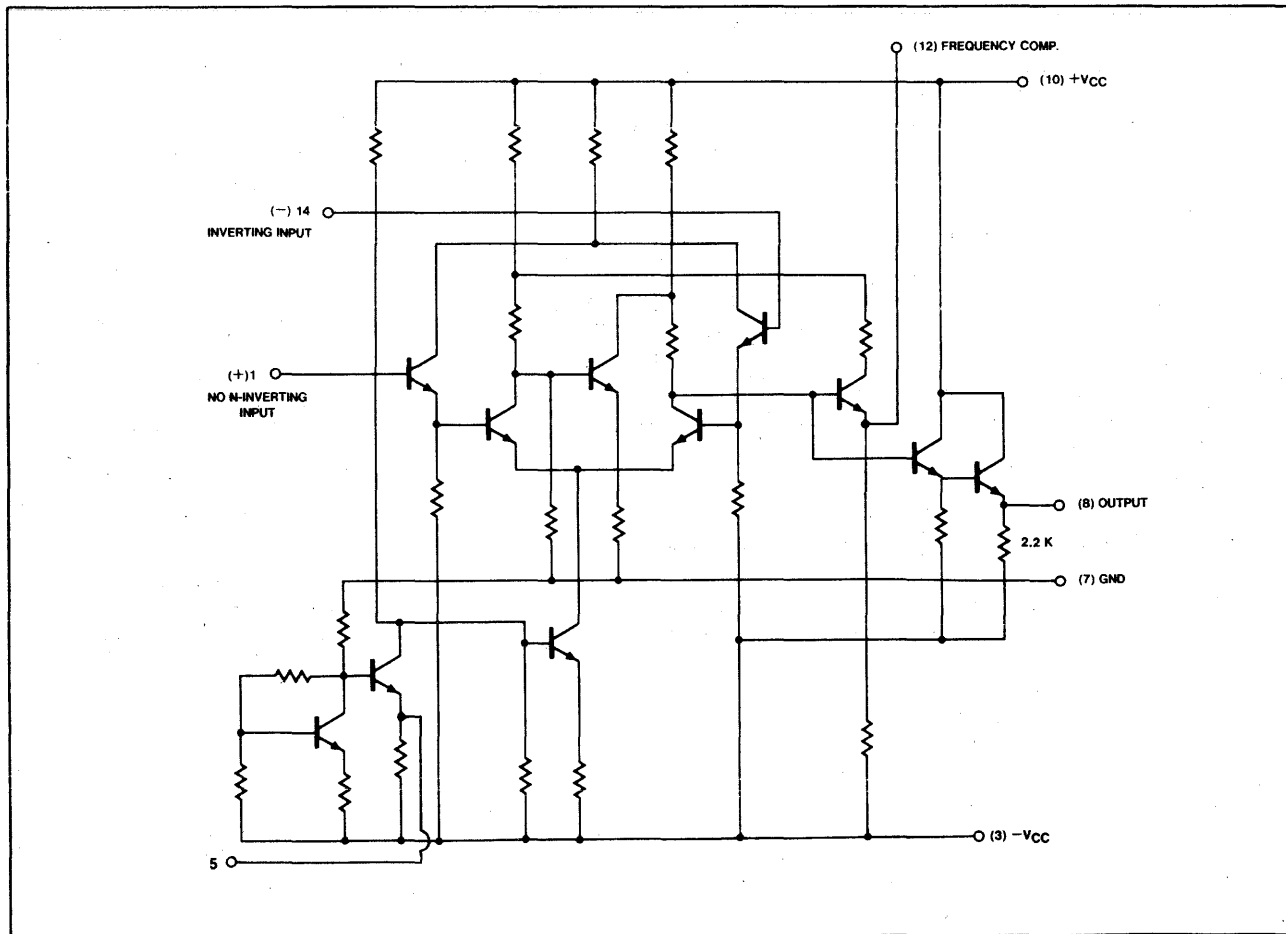
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
VCC	Supply voltage	$\pm 12$ V
PD	Internal power dissipation	550 mW
TSTG	Storage temperature range	-65 to +150 $^{\circ}$ C
TJ	Max junction temperature	150 $^{\circ}$ C
TA	Operating temperature range	
	NE	0 to 70 $^{\circ}$ C
	SE	-55 to +125 $^{\circ}$ C
	Lead temperature	300 $^{\circ}$ C

**EQUIVALENT CIRCUIT**



LINEAR

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**ULTRA HIGH FREQUENCY OPERATIONAL AMPLIFIER**

**NE/SE5539**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = \pm 8V, T_A = 25^\circ C$  unless otherwise specified

PARAMETER	TEST CONDITIONS	SE5539			NE5539			UNIT
		Min	Typ	Max	Min	Typ	Max	
V <sub>OS</sub> Input offset voltage	V <sub>O</sub> = 0V, R <sub>S</sub> = 100Ω	Over temp		2	5			mV
		T <sub>A</sub> = 25°C		2	3		2.5 5	
ΔV <sub>OS</sub> /ΔT			5			5		μV/°C
I <sub>OS</sub> Input offset current		Over temp		.1	3			μA
		T <sub>A</sub> = 25°C		.1	1		2	
ΔI <sub>OS</sub> /ΔT			.5			.5		nA/°C
I <sub>B</sub> Input bias current		Over temp		6	25			μA
		T <sub>A</sub> = 25°C		5	13		5 20	
ΔI <sub>B</sub> /ΔT			10			10		nA/°C
CMRR Common mode rejection ratio	F = 1kHz, R <sub>S</sub> = 100Ω, V <sub>CM</sub> ± 1.7V		70	80		70	80	dB
		Over temp	70	80				
R <sub>IN</sub> Input impedance			100			100		kΩ
R <sub>OUT</sub> Output impedance			10			10		Ω
V <sub>OUT</sub> Output voltage swing	R <sub>L</sub> = 150Ω to GND and 470Ω to -V <sub>CC</sub>	+Swing				+2.3	+2.7	V
		-Swing				-1.7	-2.2	
V <sub>OUT</sub> Output voltage swing	R <sub>L</sub> = 2kΩ to GND	Over temp	+Swing	+2.3	+3.0			V
			-Swing	-1.5	-2.1			
		T <sub>A</sub> = 25°C	+Swing	+2.5	+3.1			V
			-Swing	-2.0	-2.7			
I <sub>CC+</sub> Positive supply current	V <sub>O</sub> = 0, R <sub>1</sub> = ∞	Over temp		14	18			mA
		T <sub>A</sub> = 25°C		14	17		14 18	
I <sub>CC-</sub> Negative supply current	V <sub>O</sub> = 0, R <sub>1</sub> = ∞	Over temp		11	15			mA
		T <sub>A</sub> = 25°C		11	14		11 15	
PSRR Power supply rejection ratio	ΔV <sub>CC</sub> = ± 1V	Over temp		300	1000			μV/V
		T <sub>A</sub> = 25°C					200 1000	
AVOL Large signal voltage gain	V <sub>O</sub> = +2.3V, -1.7V R <sub>L</sub> = 150Ω to GND, 470Ω to -V <sub>CC</sub>					47	52 57	dB
AVOL Large signal voltage gain	V <sub>O</sub> = +2.3V, -1.7V R <sub>L</sub> = 2K to GND					47	52 57	dB
AVOL Large signal voltage gain	V <sub>O</sub> = +2.5V, -2.0V R <sub>L</sub> = 2kΩ to GND	Over temp	46		60			dB
		T <sub>A</sub> = 25°C	48	53	58			

**NOTE**

1. Differential input voltage should not exceed 0.25 volts to prevent excessive input bias current and common mode voltage 2.5 volts. These voltage limits may be exceeded if current limit is 10mA.

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**ULTRA HIGH FREQUENCY OPERATIONAL AMPLIFIER**

**NE/SE5539**

**AC ELECTRICAL CHARACTERISTICS**  $V_{CC} = \pm 8V$ ,  $R_L = 150\Omega$  to GND &  $470\Omega$  to  $-V_{CC}$  unless otherwise specified.

PARAMETER	TEST CONDITIONS	SE5539			NE5539			UNIT
		Min	Typ	Max	Min	Typ	Max	
Gain bandwidth product	$A_{CL} = 7$ $V_O = 0.1 V_{p-p}$		1200			1200		MHz
Small signal bandwidth	$A_{CL} = 2$ $R_L = 150\Omega^1$		110			110		MHz
Settling time	$A_{CL} = 2$ $R_L = 150\Omega^1$		15			15		nSec
Slew rate	$A_{CL} = 2$ $R_L = 150\Omega^1$		600			600		V/ $\mu$ Sec
Propagation delay	$A_{CL} = 2$ $R_L = 150\Omega^1$		7			7		nSec
Full power response	$A_{CL} = 2$ $R_L = 150\Omega^1$		48			48		MHz
Full power response	$A_V = 7$ , $R_L = 150\Omega$		20			20		MHz
Wide band noise (RMS)	$B_W = 5MHz$ , $R_S = 50\Omega$		4			4		nV/ $\sqrt{Hz}$

NOTE 1: External compensation.

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = \pm 6V$ ,  $T_A = 25^\circ C$  unless otherwise specified

PARAMETERS	TEST CONDITIONS	SE5539			UNIT	
		Min	Typ	Max		
$V_{OS}$ Input offset voltage	Over temp		2	5	mV	
	$T_A = 25^\circ C$		2	3		
$I_{OS}$ Input offset current	Over temp		.1	3	$\mu A$	
	$T_A = 25^\circ C$		.1	1		
$I_B$ Input bias current	Over temp		5	20	$\mu A$	
	$T_A = 25^\circ C$		4	10		
CMRR Common mode rejection ratio	$V_{CM} = \pm 1.3V$ , $R_S = 100\Omega$	70	85		dB	
$I_{CC+}$ Positive supply current	Over temp		11	14	mA	
	$T_A = 25^\circ C$		11	13		
$I_{CC-}$ Negative supply current	Over temp		8	11	mA	
	$T_A = 25^\circ C$		8	10		
PSRR Power supply rejection ratio	$\Delta V_{CC} = \pm 1V$		300	1000	$\mu V/V$	
	$T_A = 25^\circ C$					
$V_{OUT}$ Output voltage swing	$R_L = 150\Omega$ to GND and $390\Omega$ to $-V_{CC}$	Over temp	+ Swing	+ 1.4	+ 2.0	V
			- Swing	- 1.1	- 1.7	
	$T_A = 25^\circ C$	+ Swing	+ 1.5	+ 2.0		
		- Swing	- 1.4	- 1.8		

**AC ELECTRICAL CHARACTERISTICS**  $V_{CC} = \pm 6V$ ,  $R_L = 150\Omega$  to GND and  $390\Omega$  to  $-V_{CC}$  unless otherwise specified

PARAMETER	TEST CONDITIONS	SE5539			UNIT
		Min	Typ	Max	
Gain bandwidth product	$A_{CL} = 7$		700		MHz
Small signal bandwidth	$A_{CL} = 2^1$		120		MHz
Settling time	$A_{CL} = 2^1$		23		ns
Slew rate	$A_{CL} = 2^1$		330		V/ $\mu s$
Propagation delay	$A_{CL} = 2^1$		4.5		ns
Full power response	$A_{CL} = 2^1$		20		MHz

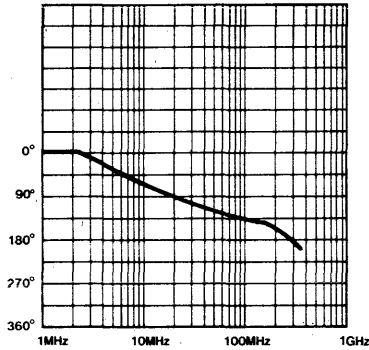
NOTE 1: External compensation.

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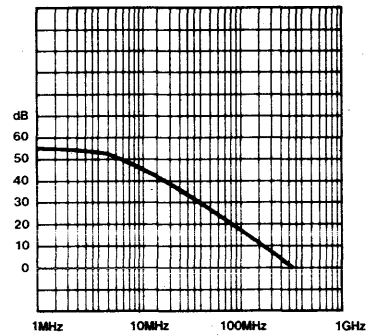
ULTRA HIGH FREQUENCY OPERATIONAL AMPLIFIER

NE/SE5539

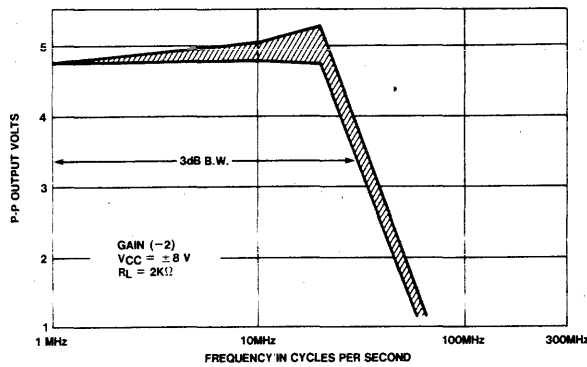
NE5539 OPEN LOOP PHASE



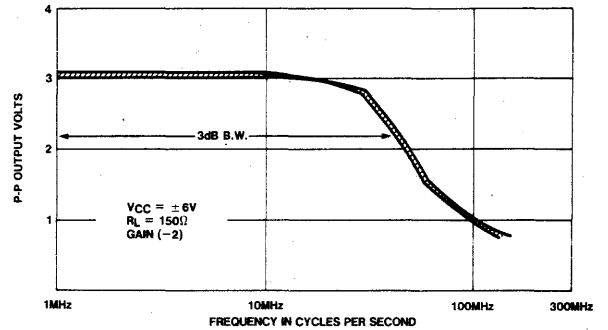
NE5539 OPEN LOOP GAIN



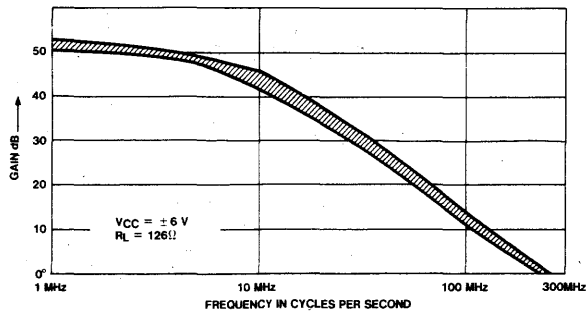
POWER BANDWIDTH (SE)



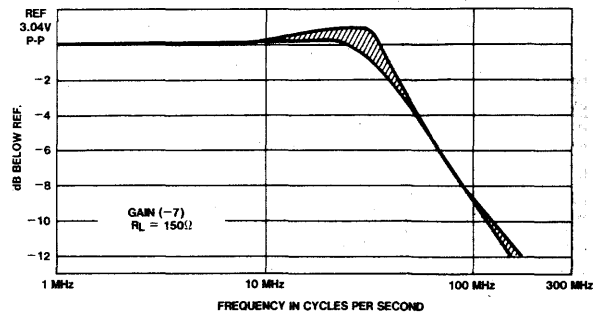
POWER BANDWIDTH (NE)



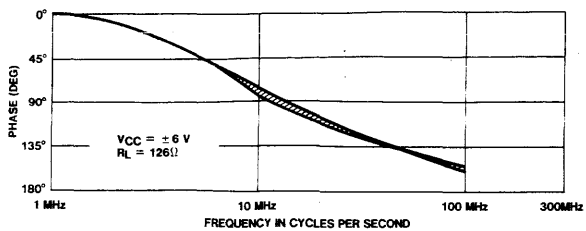
SE5539 OPEN LOOP GAIN vs FREQUENCY



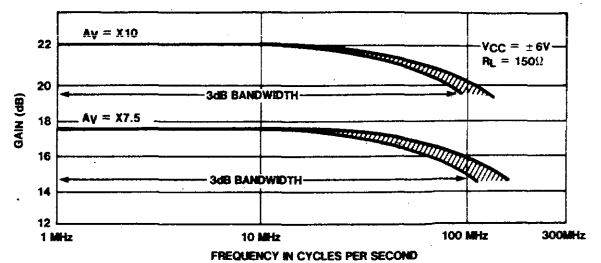
POWER BANDWIDTH



SE5539 OPEN LOOP PHASE vs FREQUENCY



GAIN BANDWIDTH PRODUCT vs FREQUENCY



NOTE

Indicates typical distribution  $-55^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$

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ULTRA HIGH FREQUENCY OPERATIONAL AMPLIFIER

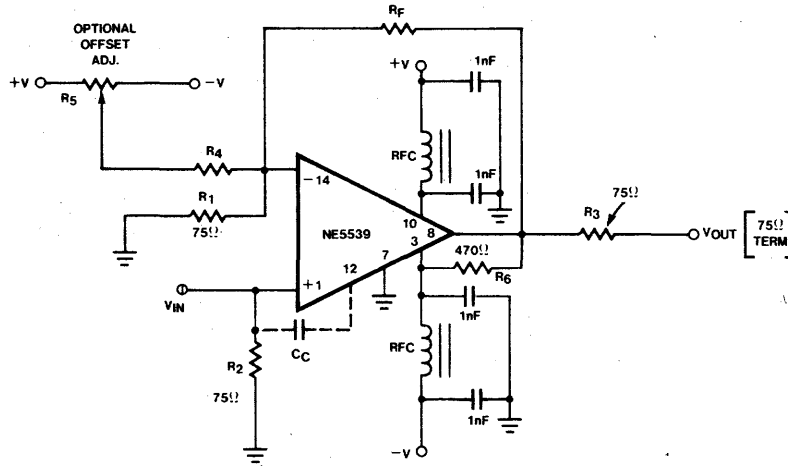
NE/SE5539

CIRCUIT LAYOUT CONSIDERATIONS

As may be expected for an ultra-high frequency, wide gain bandwidth amplifier, the physical circuit layout is extremely

critical. Breadboarding is not recommended. A double-sided copper clad printed circuit board will result in more

favorable system operation. An example utilizing a 28dB non-inverting amp is shown in Figure 1.



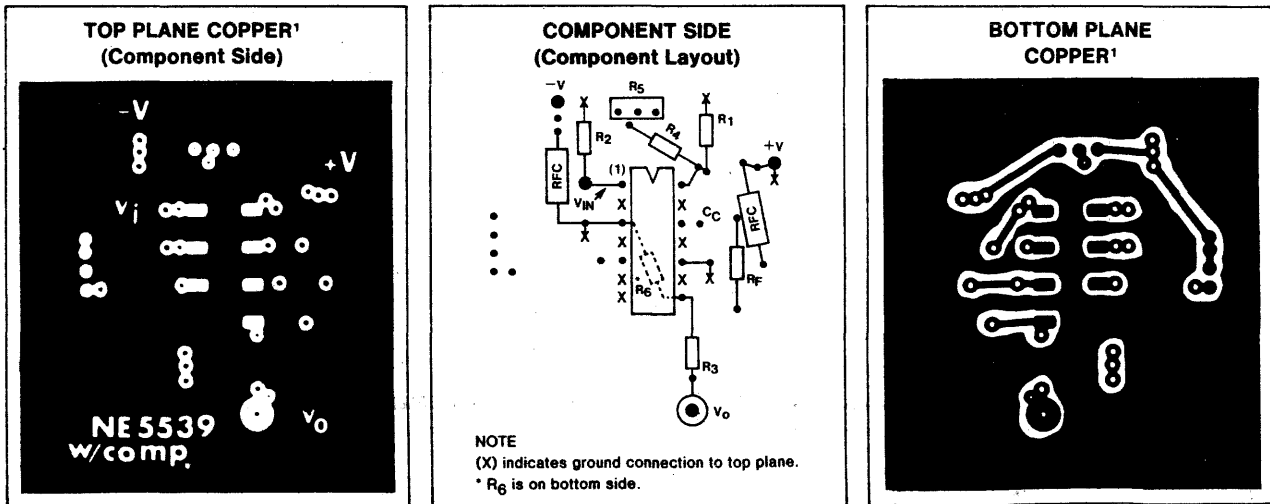
R<sub>1</sub> = 75Ω 5% CARBON  
 R<sub>2</sub> = 75Ω 5% CARBON  
 R<sub>3</sub> = 75Ω 5% CARBON  
 R<sub>4</sub> = 36K 5% CARBON

R<sub>5</sub> = 20K TRIMPOT (CERMET)  
 R<sub>f</sub> = 1.5K (28dB GAIN)  
 R<sub>6</sub> = 470Ω 5% CARBON

RFC 3T # 26 BUSSWIRE ON FERROXCLUBE VK 200 09 '3B CORE  
 BYPASS CAPACITORS 1nF CERAMIC (MEPCO OR EQUIV.)

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NOTE 1: Bond edges of top and bottom ground plane copper.

Figure 1. 28dB Non-Inverting Amp Sample P.C. Layout

**VIDEO AMPLIFIER**

**NE5592**

**Preliminary**

**DESCRIPTION**

The NE5592 is a dual monolithic, two stage, differential output, wideband video amplifier. It offers fixed gain of 400 without external components or adjustable gains from 400 to 0 with one external resistor. The input stage has been designed so that with the addition of a few external reactive elements between the gain select terminals, the circuit can function as a high pass, low pass, or band pass filter. This feature makes the circuit ideal for use as a video or pulse amplifier in communications, magnetic memories, display, video recorder systems, and floppy disk head amplifiers.

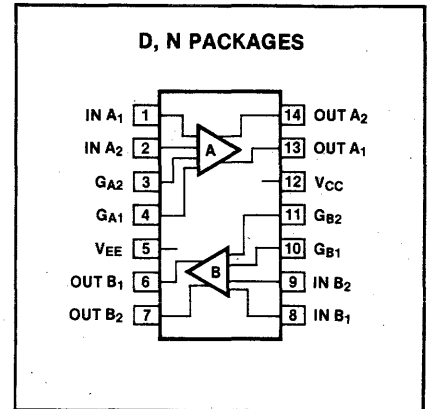
**FEATURES**

- 120MHz bandwidth
- Adjustable gains from 0 to 400
- Adjustable pass band
- No frequency compensation required
- Wave shaping with minimal external components

**APPLICATIONS**

- Floppy disk head amplifier
- Video amplifier
- Pulse amplifier in communications
- Magnetic memory
- Video recorder systems

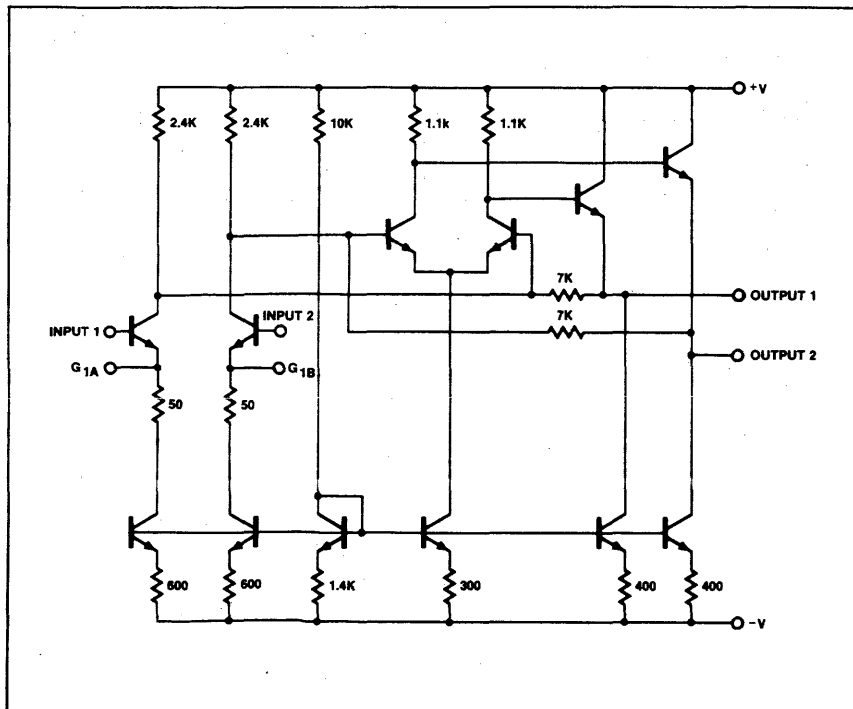
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**  $T_A = 25^\circ\text{C}$  unless otherwise specified.

PARAMETER	RATING	UNIT
Supply voltage	$\pm 8$	V
Differential input voltage	$\pm 5$	V
Common mode Input voltage	$\pm 6$	V
Output current	10	mA
Operating temperature range	0 to +70	$^\circ\text{C}$
NE5592	-65 to +150	$^\circ\text{C}$
Power dissipation	500	mW

**EQUIVALENT CIRCUIT**



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**VIDEO AMPLIFIER**

**NE5592**

**Preliminary**

**DC ELECTRICAL CHARACTERISTICS**  $T_A = +25^\circ\text{C}$ ,  $V_{SS} = \pm 6\text{V}$ ,  $V_{CM} = 0$  unless otherwise specified.  
Recommended operating supply voltage  $V_S = \pm 6.0\text{V}$ .

PARAMETER	TEST CONDITIONS	NE5592			UNITS
		Min	Typ	Max	
Differential voltage gain Gain 1 <sup>1</sup>	$R_L = 2\text{k}\Omega$ , $V_{OUT} = 3\text{V p-p}$	250	400	600	V/V
Bandwidth Gain 1 <sup>1</sup>			40		MHz
Rise Time Gain 1 <sup>1</sup>	$V_{OUT} = 1\text{V p-p}$		10.5		ns
Propagation delay Gain 1 <sup>1</sup>	$V_{OUT} = 1\text{V p-p}$		7.5		ns
Input resistance Gain 1 <sup>1</sup>			4.0		k $\Omega$
Input capacitance <sup>1</sup>			2.0		pF
Input offset current			0.4	5.0	$\mu\text{A}$
Input bias current			9.0	30	$\mu\text{A}$
Input noise voltage	BW 1kHz to 10MHz		12		$\mu\text{Vrms}$
Input voltage range				$\pm 1.0$	V
Common mode rejection ratio Gain 1 <sup>1</sup>	$V_{CM} \pm 1\text{V}$ , $F < 100\text{kHz}$	60	86		dB
Gain 1 <sup>1</sup>	$V_{CM} \pm 1\text{V}$ , $F = 5\text{MHz}$		60		dB
Supply voltage rejection ratio Gain 1 <sup>1</sup>	$\Delta V_S = \pm 0.5\text{V}$	50	70		dB
Channel separation	$V_{OUT} = 3\text{V}_{pp}$ ; $F = 100\text{K}$ (input referenced)		85		dB
Output offset voltage Gain 1 <sup>1</sup>	$R_L = \infty$			1.5	V
Gain 2 <sup>2</sup>	$R_L = \infty$		0.35	0.75	V
Output common mode voltage	$R_L = \infty$	2.4	2.9	3.4	V
Output differential voltage swing	$R_L = 2\text{k}$	3.0	4.0		V
Output resistance			20		$\Omega$
Power supply current	$R_L = \infty$		18	24	mA
<b>THE FOLLOWING SPECS APPLY OVER TEMPERATURE</b> $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$					
Differential voltage gain Gain 1 <sup>1</sup>	$R_L = 2\text{k}\Omega$ , $V_{OUT} = 3\text{V p-p}$	250		600	V/V
Input resistance Gain 1 <sup>1</sup>					k $\Omega$
Input offset current				6.0	$\mu\text{A}$
Input bias current				40	$\mu\text{A}$
Input voltage range		$\pm 1.0$			V
Common mode rejection ratio Gain 1 <sup>1</sup>	$V_{CM} \pm 1\text{V}$ , $F < 100\text{kHz}$	50			dB
Supply voltage rejection ratio Gain 1 <sup>1</sup>	$\Delta V_S = \pm 0.5\text{V}$	50			dB
Channel separation	$V_{OUT} = 3\text{V}_{pp}$ ; $F = 100\text{K}$ (input referenced)		75		dB
Output offset voltage Gain 1	$R_L = \infty$			1.5	V
Gain 2 <sup>2</sup>	$R_L = \infty$			1.0	V
Output differential voltage swing	$R_L = 2\text{k}$	2.8			V
Power supply current	$R_L = \infty$			27	mA

NOTES:  
1. Gain select pins connected together.  
2. All gain select pins open.

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# Si Telecommunications Integrated Circuit Capabilities

Silicon Systems offers a broad line of standard telecommunications circuits aimed at providing cost-effective solutions for common customer application problems. At the heart of SSI's efforts in the communications market is its pioneering work with CMOS switched capacitor filters. Our early success with the DTMF receiver has enabled us to develop a family of chips utilizing the switched capacitor filter technology.

As a trendsetter in the field, Silicon Systems is leading the way towards a whole new era of VLSI circuits for telecommunications. Our broad selection of DTMF receivers demonstrates not only technological leadership in our own semiconductor field but also our capability to anticipate the growing needs of the fast-paced telecommunications marketplace.

## PROCESSES

Silicon Systems offers circuits in junction-isolated, bipolar, single and double-layer metal. Plus, SSI has a CMOS capability that includes not only a metal-gate process but also a silicon-gate process that produces circuits packed with more functions in a smaller size for high-speed, low-power performance. These are the most popular and reliable processes in the two basic technologies, and SSI's advanced ultra-clean wafer fab produces higher yields than ordinary facilities.

## PRODUCT QUALITY

Silicon Systems has made a major investment in product test and in-line quality control equipment. For example, a state-of-the-art LTX CP80 is used for functional and parametric testing of sophisticated analog, digital, and combination A/D circuits. In this way, SSI is dedicated to the delivery of complex VLSI circuits to meet the incoming quality level you require.

Here are a few completed circuits that demonstrate our broad telecommunications IC capability:

### BIPOLAR

Integrated Circuit Function	Application
Audio System Receiver	Telephone Answering Machine
VHF/UHF Gain Mixer	Radio Receiver
Pulse Width Modulator	Switching Power Supply
Controller	Home Appliance
Digital Receiver	Remote Control
PCM Encoder/Decoder	Telecom System
Digital Correlator/Integrator	Radio Telescope

### MOS

Integrated Circuit Function	Application
DTMF Receiver	*Decodes Touch-Tone® Telephone Signals
300 Baud Modem	Data Transmission
1200/2400 Baud Receiver	FSK/PSK Modem
Error Corrector	Military Radio
Remote Transmitter	Telephone Answering Machine
Phoneme Based Speech Synthesizer	Text-to-Speech
Display Timing Generator	TV Sets
Video Processor	Infrared Video System
16 Channel Switching Matrix	Bank Communications System
Digital Loop Detector	Traffic Signal Control
Programmable Digital Receiver	Home Appliance Remote Control
Vocal Tract System	Speech Synthesis

## SSI STANDARD PRODUCTS TELECOMMUNICATIONS CIRCUITS

Part Number	Circuit Type	Package
<b>DTMF's Receivers/Filters</b>		
SSI 201	12V, DTMF Receiver	22 DIP
SSI 202	5V, DTMF Receiver	18 DIP
SSI 203	5V, DTMF Receiver (w/early detect)	18 DIP
SSI 5601	5V/12V, Dial Tone Reject Filter	14 DIP
<b>Modem</b>		
SSI 180	1200 Baud FSK Modem	16 DIP
<b>Speech Synthesizer</b>		
SSI SC-01	VOTRAX SC-01 Compatible	22 DIP
SSI 263	VOTRAX SC-02 Compatible	24 DIP
<b>Switched Capacitor Filter Array</b>		
SCA-6	CMOS Semicustom SCF Array ( 6 filter sections)	8-28 DIP
SCA-12	CMOS Semicustom SCF Array (12 filter sections)	24-40 DIP

**Si Silicon Systems  
incorporated**

14351 Myford Road, Tustin, California 92680 (714) 731-7110, TWX 910-595-2809



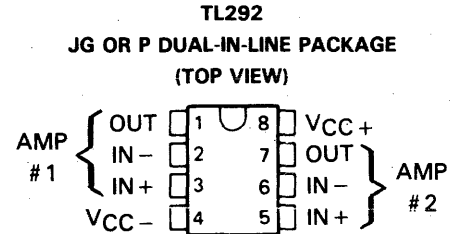
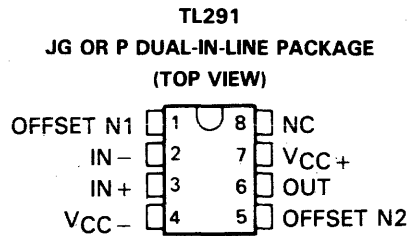
**Don't miss  
the Sections on  
Semicustom ICs**

**Digital Gate Arrays,  
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Linear Arrays Are  
All Covered**

**Vol. II, page 4425**



- Small-Signal Unity-Gain Bandwidth . . . 20 MHz Typ
- Noninverting Slew Rate . . . 50 V/ $\mu$ s Typ (Unity-Gain Follower)
- Internal Frequency Compensation
- Full-Power Bandwidth at  $V_{OPP} = 20$  V . . . 400 kHz Typ
- Open-Loop Gain at Full-Power Bandwidth,  $V_{OPP} = 20$  V . . . 34 dB Typ
- Output Short-Circuit Protection
- TL291 Has Offset Null Capability
- Pinout is Same as Standard General Purpose Operational Amplifiers



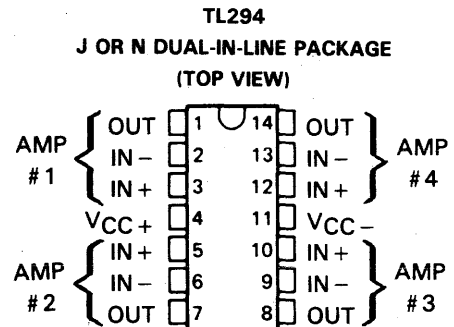
**description**

These devices are high-speed operational amplifiers designed for applications requiring wide bandwidth and a fast slew rate. These monolithic circuits incorporate new high-frequency P-N-P transistors that eliminate the need for large feed-forward capacitors required in previous moderately high-frequency designs to pass the signal around slow lateral P-N-P stages.

These operational amplifiers have a typical full-power bandwidth of 400 kilohertz for a 20-volt peak-to-peak output swing. because of the higher 20-megahertz unity-gain bandwidth, the typical open-loop gain at the 400-kilohertz full-power bandwidth is a very respectable 34 decibels.

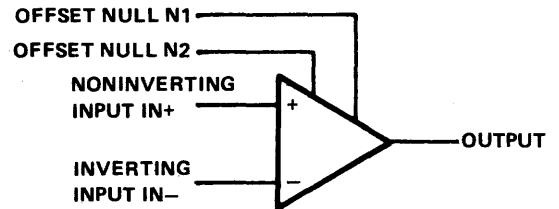
The TL291 single-channel operational amplifier pinout includes offset nulling, which is easily accomplished by connecting a potentiometer across the offset null pins with the wiper connected to the  $V_{CC-}$  pin.

The TL291M, TL292M, and TL294M will be characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The TL291C, TL292C, and TL294C will be characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .



NC—No internal connection

**symbol (each amplifier)**



**PRODUCT PREVIEW**

This document contains information on a product under development. Texas Instruments reserves the right to change or discontinue this product without notice.

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INSTRUMENTS**

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SLOS004

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**ALABAMA:** Huntsville, 500 Wynn Drive, Suite 514, Huntsville, AL 35805, (205) 837-7530.

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**CALIFORNIA:** El Segundo, 831 S. Douglas St., El Segundo, CA 90245, (213) 973-2571; Irvine, 17891 Cartwright Rd., Irvine, CA 92714, (714) 660-1200; Sacramento, 1900 Point West Way, Suite 171, Sacramento, CA 95815, (916) 929-1521; San Diego, 4333 View Ridge Ave., Suite B, San Diego, CA 92123, (714) 278-9600; Santa Clara, 5353 Betsy Ross Dr., Santa Clara, CA 95054, (408) 980-9000; Woodland Hills, 21220 Erwin St., Woodland Hills, CA 91367, (213) 704-7759.

**COLORADO:** Denver, 9725 E. Hampden St., Suite 301, Denver, CO 80231, (303) 695-2800.

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**INDIANA:** Ft. Wayne, 2020 Inwood Dr., Ft. Wayne, IN 46805, (219) 424-5174; Indianapolis, 2346 S. Lynhurst, Suite J-400, Indianapolis, IN 46241, (317) 248-8555.

**IOWA:** Cedar Rapids, 373 Collins Rd. NE, Suite 200, Cedar Rapids, IA 52402, (319) 395-9550.

**MARYLAND:** Baltimore, 1 Rutherford Pl., 7133 Rutherford Rd., Baltimore, MD 21207, (301) 944-8600.

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**OREGON:** Beaverton, 6700 SW 105th St., Suite 110, Beaverton, OR 97005, (503) 643-6758.

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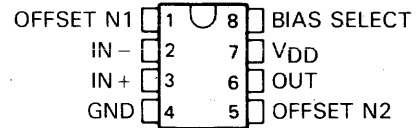
BD

- **Wide range of Supply Voltages:**  
1 V to 16 V (TLC251)  
4 V to 16 V (TLC271)
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Includes the Negative Rail**
- **Low Noise . . . 30 nV/√Hz Typ at 1 kHz (High-Bias)**

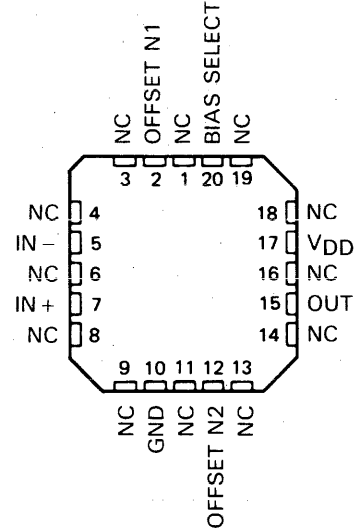
**description**

The TLC251 and TLC271 series are low-cost, low-power programmable operational amplifiers designed to operate with single or dual supplies. Unlike traditional metal-gate CMOS op amps, these devices utilize the Texas Instruments silicon gate LinCMOST™ process, giving them stable input offset voltages without sacrificing the advantages of metal-gate CMOS. This series of parts is available in selected grades of input offset voltage and can be nulled with one external potentiometer. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this family is ideally suited for battery-powered or energy-conserving applications. A bias-select pin can be used to program one of three ac performance and power-dissipation levels to suit the application. The TLC251 offers the same operation as the TLC271, but also features guaranteed operation down to a 1 V supply. Both devices are stable at unity gain.

**D, JG, OR P DUAL-IN-LINE PACKAGE  
(TOP VIEW)**

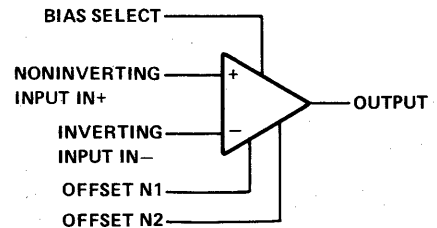


**FH OR FK PACKAGE  
(TOP VIEW)**



NC No internal connection

**symbol**



**TEMPERATURE RANGES AND PACKAGES**

SERIES	TEMPERATURE RANGE	PACKAGES
TLC251C Types	0°C to 70°C	JG, P, D
TLC271C Types	0°C to 70°C	JG, P, D
TLC271I Types	-40°C to 85°C	JG, P, D
TLC271M Types	-55°C to 125°C	JG, FH, FK

**DEVICE FEATURES**

PARAMETER	LOW BIAS	MEDIUM BIAS	HIGH BIAS
Supply current (Typ)	10 μA	150 μA	1000 μA
Slew rate (Typ)	0.04 V/μs	0.6 V/μs	4.5 V/μs
Input offset voltage (Max)			
... Standard types	10 mV	10 mV	10 mV
... A-suffix types	5 mV	5 mV	5 mV
... B-suffix types	2 mV	2 mV	2 mV
Offset voltage change (Typ)	0.1 μV/month† 0.7 μV/°C	0.1 μV/month† 2 μV/°C	0.1 μV/month† 5 μV/°C
Input bias current (Typ)	1 pA	1 pA	1 pA
Input offset current (Typ)	1 pA	1 pA	1 pA

†The long-term drift value applies after the first month.

**ADVANCE INFORMATION**

This document contains information on a new product. Specifications are subject to change without notice.

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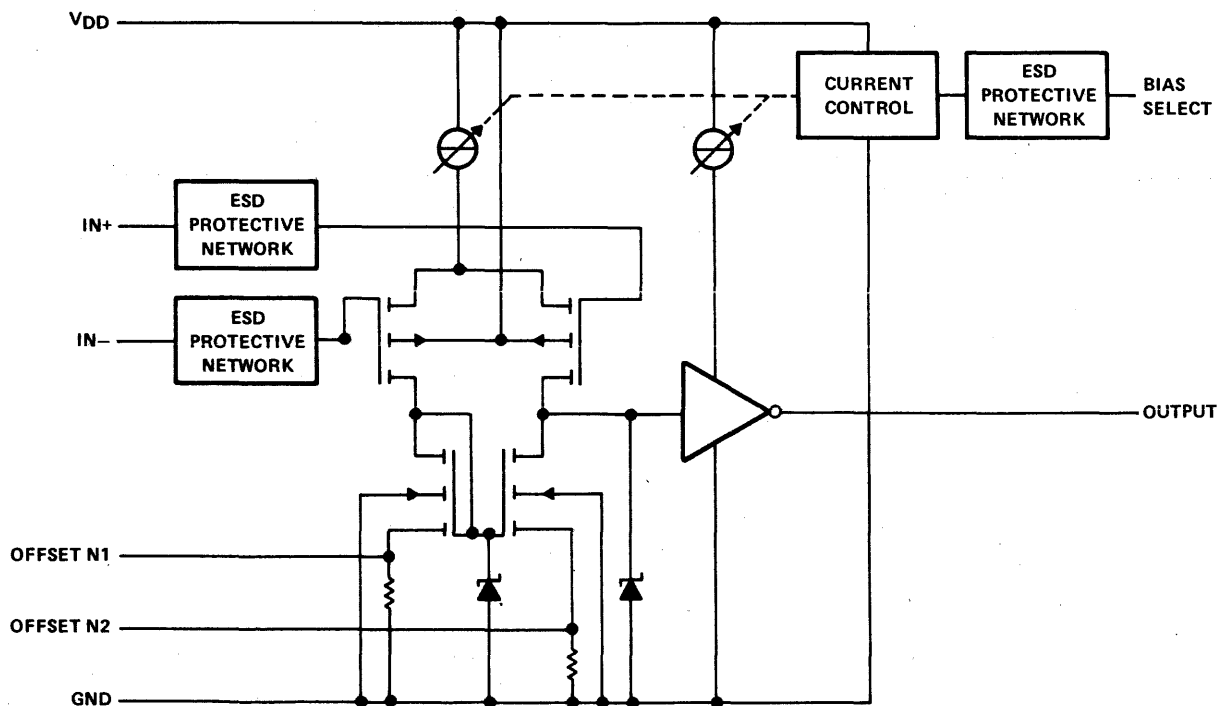
# TYPES TLC251, TLC251A, TLC251B, TLC271, TLC271A, TLC271B PROGRAMMABLE LOW-POWER LinCMOS™ AMPLIFIERS

These devices have internal Electrostatic Discharge (ESD) protection circuits which will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883B Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC251 and TLC271 series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS operational amplifiers without the power penalties of traditional bipolar devices. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC271. Remote and inaccessible equipment applications are possible using the low-voltage and low-power capabilities of the TLC251. In addition, by driving the bias-select input with a logic signal from a microprocessor, these operational amplifiers can have software-controlled performance and power consumption. The TLC251 is well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications.

Devices having an "M" suffix are characterized for operation over the temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ , those with an "I" suffix are characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , and those with a "C" suffix are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## schematic



TEXAS  
INSTRUMENTS

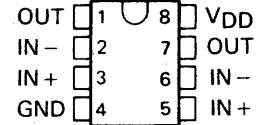
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- **Wide range of Supply Voltages:**  
1 V to 16 V (TLC252 types)  
4 V to 16 V (TLC272 types)
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Includes the Negative Rail**
- **Low Noise . . . 30 nV $\sqrt{\text{Hz}}$  Typ at f = 1 kHz (High-Bias Versions)**

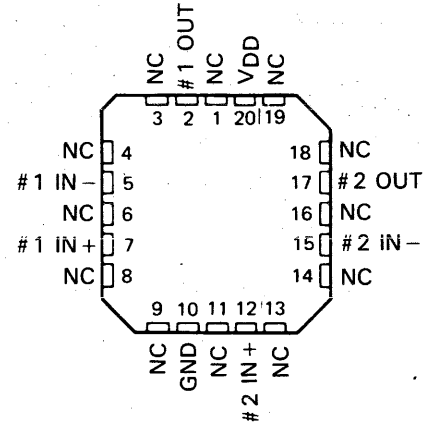
**description**

The TLC252 and TLC272 series are low-cost, low-power dual operational amplifiers designed to operate with single or dual supplies. These devices utilize the Texas Instruments silicon gate LinCMOSTM process, giving them stable input offset voltages that are available in selected grades of 2, 5 or 10 mV maximum, very high input impedances, and extremely low input offset and bias currents. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this series is ideally suited for battery-powered or energy-conserving applications. The TLC252 types offer guaranteed operation down to a 1-V supply. All devices are unity-gain stable and have excellent noise characteristics.

**D, JG, OR P DUAL-IN-LINE PACKAGE  
(TOP VIEW)**

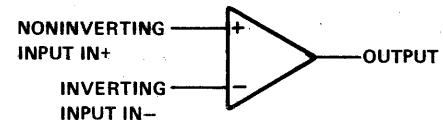


**FH OR FK PACKAGE  
(TOP VIEW)**



NC—No internal connection

**symbol (each amplifier)**



**DEVICE FEATURES**

PARAMETER	TLC25L2 TLC27L2 (LOW BIAS)	TLC25M2 TLC27M2 (MEDIUM BIAS)	TLC252 TLC272 (HIGH BIAS)
Supply current (Typ)	20 $\mu\text{A}$	300 $\mu\text{A}$	2000 $\mu\text{A}$
Slew rate (Typ)	0.04 V/ $\mu\text{s}$	0.6 V/ $\mu\text{s}$	4.5 V/ $\mu\text{s}$
Input offset voltage (Max)			
... Standard types	10 mV	10 mV	10 mV
... A-suffix types	5 mV	5 mV	5 mV
... B-suffix types	2 mV	2 mV	2 mV
Offset voltage change (Typ)	0.1 $\mu\text{V}/\text{month}^\dagger$ 0.7 $\mu\text{V}/^\circ\text{C}$	0.1 $\mu\text{V}/\text{month}^\dagger$ 2 $\mu\text{V}/^\circ\text{C}$	0.1 $\mu\text{V}/\text{month}^\dagger$ 5 $\mu\text{V}/^\circ\text{C}$
Input bias current (Typ)	1 pA	1 pA	1 pA
Input offset current (Typ)	1 pA	1 pA	1 pA

$^\dagger$ The offset voltage drift applies after the first month only.

**TEMPERATURE RANGES AND PACKAGES**

SERIES	TEMPERATURE RANGE	PACKAGES
TLC252C Types	0°C to 70°C	JG, P, D
TLC272C Types	0°C to 70°C	JG, P, D
TLC272I Types	-40°C to 85°C	JG, P, D
TLC272M Types	-55°C to 125°C	JG, FH, FK

**ADVANCE INFORMATION**

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**TEXAS  
INSTRUMENTS**

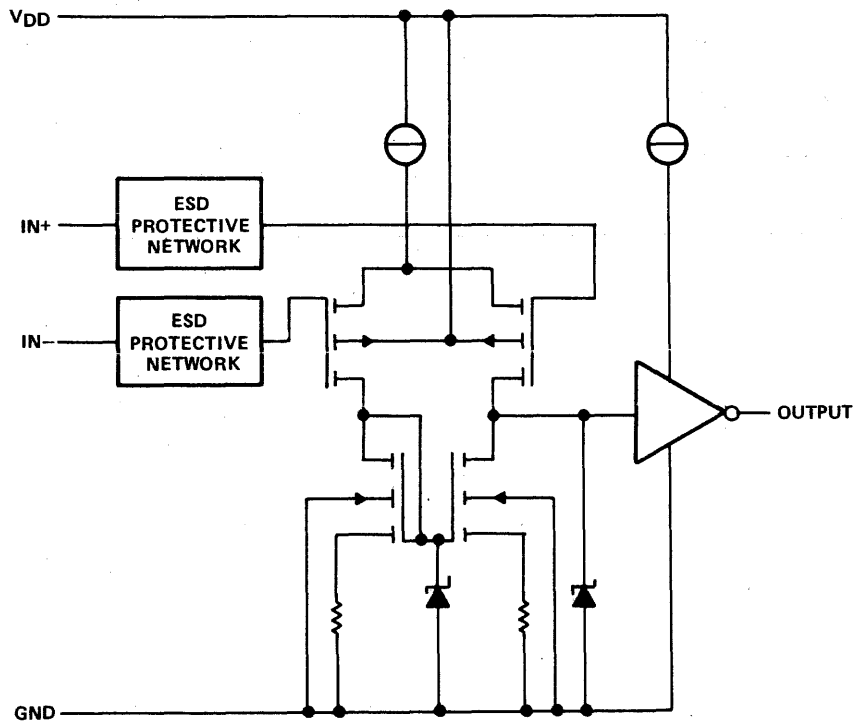
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# TYPES TLC252, TLC25L2, TLC25M2, TLC272, TLC27L2, TLC27M2 LinCMOST™ DUAL OPERATIONAL AMPLIFIERS

These devices have internal Electrostatic Discharge (ESD) protection circuits which will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883B Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC252 and TLC272 series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOST™ operational amplifiers without the power penalties of traditional bipolar devices. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC252 and TLC272 series. Remote and inaccessible equipment applications are possible using the low-voltage and low-power capabilities of the TLC252. The TLC252 types are well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications. This series includes devices that are characterized for commercial, industrial, and military temperature ranges and are available in 8-pin plastic and ceramic dual-in-line (DIP) packages, small outline (D) package, and chip carrier (FH, FK) packages.

schematic (each amplifier)



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# TYPES TLC254, TLC25L4, TLC25M4, TLC274, TLC27L4, TLC27M4 LinCMOSTM QUAD OPERATIONAL AMPLIFIERS

D2753, JUNE 1983—REVISED OCTOBER 1983

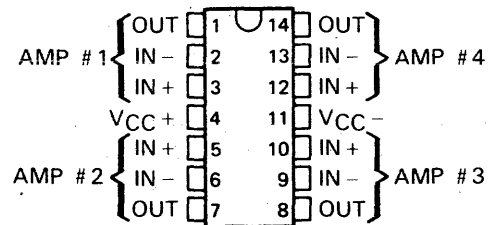
- **Wide range of Supply Voltages:**  
1 V to 16 V (TLC254 types)  
4 V to 16 V (TLC274 types)
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Includes the Negative Rail**
- **Low Noise . . . 30 nV $\sqrt{\text{Hz}}$  Typ at f = 1 kHz (High-Bias Versions)**

## description

The TLC254 and TLC274 series are low-cost, low-power dual operational amplifiers designed to operate with single or dual supplies. These devices utilize the Texas Instruments silicon gate LinCMOSTM process, giving them stable input offset voltages that are available in selected grades of 2, 5 or 10 mV maximum, very high input impedances, and extremely low input offset and bias currents. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this series is ideally suited for battery-powered or energy-conserving applications. The TLC254 types offer guaranteed operation down to a 1-V supply. All devices are unity-gain stable and have excellent noise characteristics.

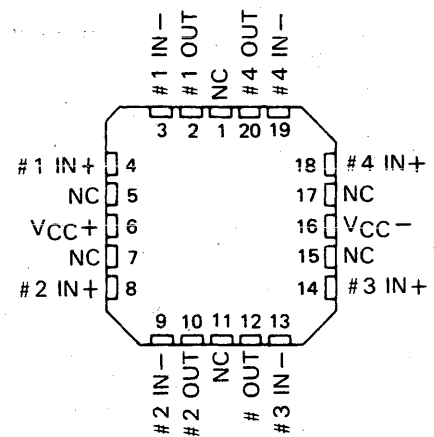
## D, J, OR N DUAL IN-LINE-PACKAGE

(TOP VIEW)



## FH OR FK PACKAGE

(TOP VIEW)

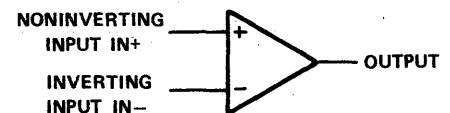


NC—No internal connection

## TEMPERATURE RANGES AND PACKAGES

DEVICE TYPES	TEMPERATURE RANGE	PACKAGES
TLC254C	0°C to 70°	J, N, D
TLC274C	0°C to 70°C	J, N, D
TLC274I	-40°C to 85°C	J, N, D
TLC274M	-55°C to 125°C	J, FH, FK

## symbol (each amplifier)



## DEVICE FEATURES

PARAMETER	TLC25L4 TLC27L4 (LOW BIAS)	TLC25M4 TLC27M4 (MEDIUM BIAS)	TLC254 TLC274 (HIGH BIAS)
Supply current (Typ)	40 $\mu\text{A}$	600 $\mu\text{A}$	4000 $\mu\text{A}$
Slew rate (Typ)	0.04 V/ $\mu\text{s}$	0.6 V/ $\mu\text{s}$	4.5 V/ $\mu\text{s}$
Input offset voltage (Max)			
... Standard types	10 mV	10 mV	10 mV
... A-suffix types	5 mV	5 mV	5 mV
... B-suffix types	2 mV	2 mV	2 mV
Offset voltage change (Typ)	0.1 $\mu\text{V}/\text{month}^\dagger$ 0.7 $\mu\text{V}/^\circ\text{C}$	0.1 $\mu\text{V}/\text{month}^\dagger$ 2 $\mu\text{V}/^\circ\text{C}$	0.1 $\mu\text{V}/\text{month}^\dagger$ 5 $\mu\text{V}/^\circ\text{C}$
Input bias current (Typ)	1 pA	1 pA	1 pA
Input offset current (Typ)	1 pA	1 pA	1 pA

$^\dagger$ The long-term drift value applies after the first month.

## ADVANCE INFORMATION

This document contains information on a new product. Specifications are subject to change without notice.

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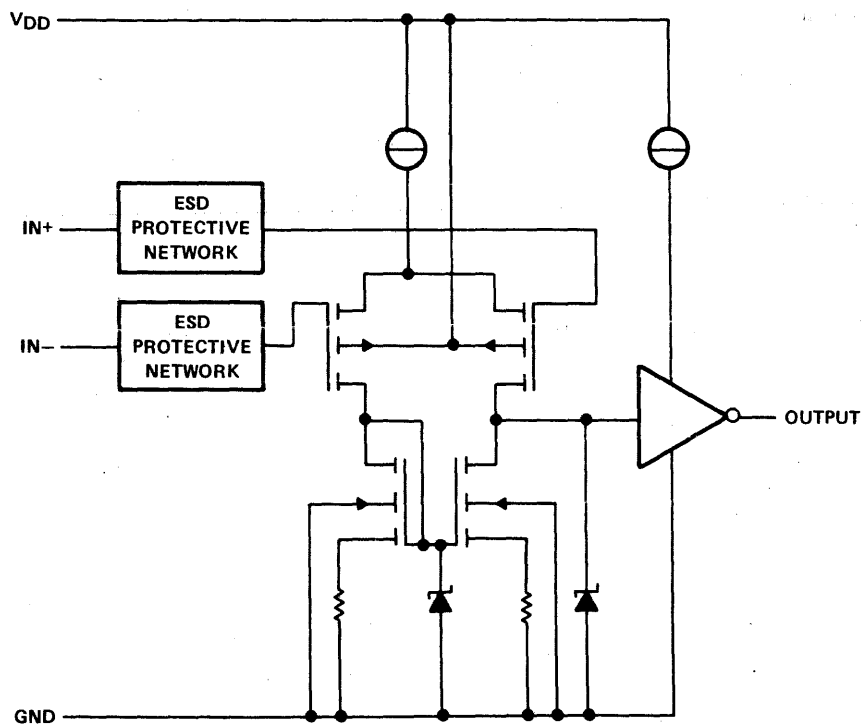
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# TYPES TLC254, TLC25L4, TLC25M4, TLC274, TLC27L4, TLC27M4 LinCMOST™ QUAD OPERATIONAL AMPLIFIERS

These devices have internal Electrostatic Discharge (ESD) protection circuits which will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883B Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC254 and TLC274 series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOST™ operational amplifiers without the power penalties of traditional bipolar devices. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC254 and TLC274 series. Remote and inaccessible equipment applications are possible using the low-voltage and low-power capabilities of the TLC254. The TLC254 types are well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications. This series includes devices that are characterized for commercial, industrial, and military-temperature ranges and are available in 14-pin plastic and ceramic dual-in-line (DIP) packages, small outline (D) package, and chip carrier (FH, FK) packages.

schematic (each amplifier)

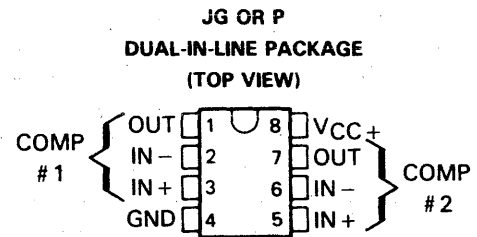


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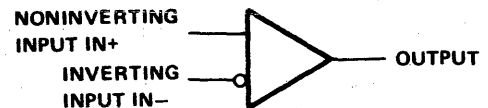
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- Single- or Dual-Supply Operation
- Wide Range of Supply Voltages  
2 to 18 Volts
- Very Low Supply Current Drain  
0.2 mA Typ
- Fast Response Time . . . 200 ns Typ for  
TTL-Level Input Step
- Built-In ESD Protection
- High Input Impedance . . .  $10^{12} \Omega$  Typ
- Extremely Low Input Bias Current  
1 pA Typ
- Ultra-Stable Low Input Offset Voltage
- Common-Mode Input Voltage Range  
Includes Ground
- Output Compatible with TTL, MOS, and  
CMOS



symbol (each comparator)



**description**

This device is fabricated using LinCMOS™ technology and consists of two independent voltage comparators designed to operate from a single power supply. Operation from dual supplies is also possible so long as the difference between the two supplies is 2 to 18 volts. Each of these devices features extremely high input impedance (typically greater than  $10^{12}$  ohms) allowing direct interfacing with high-impedance sources. The outputs are n-channel open-drain configurations, and can be connected to achieve positive-logic wired-AND relationships.

These devices have internal Electrostatic Discharge (ESD) protection circuits which will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883B Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

The TLC372M is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The TLC372C is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

**PRODUCT PREVIEW**

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# TYPES TLC372M, TLC372C

## DUAL LinCMOSTM DIFFERENTIAL COMPARATORS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TLC372M	TLC372C	UNIT
Supply voltage (see Note 1)	18	18	V
Differential input voltage (see Note 2)	± 18	± 18	V
Input voltage, $V_I$	18	18	V
Output voltage, $V_O$	18	18	V
Output current, $I_O$	20	20	mA
Duration of output short-circuit to ground (see Note 3)	unlimited	unlimited	
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 4)	900	900	mW
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.  
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.  
 3. Short circuits from outputs to  $V_{CC}$  can cause excessive heating and eventual destruction.  
 4. For operation above 25°C free-air temperature, refer to Dissipation Derating Curves, Section 2. In the J package, the TLC372C chips are glass-mounted; the TLC372M chips are alloy-mounted.

electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC372M			TLC372C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = V_{ICR}$ min, See Note 5	25°C	2	10	2	10	mV	
		Full range		12		12		
$I_{IO}$ Input offset current	See Note 5	25°C	1		1		pA	
		Full range		10		0.3	nA	
$I_{IB}$ Input bias current		25°C	1				pA	
		Full range		20		0.6	nA	
$V_{ICR}$ Common-mode input voltage range		25°C	0 to $V_{CC} - 1.5$		0 to $V_{CC} - 1.5$		V	
		Full range	0 to $V_{CC} - 2$		0 to $V_{CC} - 2$			
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$ , $R_L \geq 15\text{ k}\Omega$ to $V_{CC}$	25°C	200		200		v/mV	
$I_{OH}$ High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C	0.1		0.1	nA	
		$V_{OH} = 15\text{ V}$	Full range		1		$\mu\text{A}$	
$V_{OL}$ Low-level output voltage	$V_{ID} = 1\text{ V}$ , $I_{OL} = 4\text{ mA}$	25°C	150	400	150	400	mV	
		Full range		700		700		
$I_{OL}$ Low-level output current	$V_{ID} = 1\text{ V}$ , $V_{OL} = 1.5\text{ V}$	25°C	6	16	6	16	mA	
$I_{CC}$ Supply current (two comparators)	$V_{ID} = -1\text{ V}$ , No load	25°C	0.2		0.2		mA	

†All characteristics are measured with zero common-mode input voltage unless otherwise specified.

NOTE 5: The offset voltages and offset currents given are the maximum values required to drive the output up to 4 V or down to 400 mV with a pull-up resistor of 2.5 k $\Omega$  to  $V_{CC}$ . Thus, these parameters actually define an error band and take into account the worst-case effects of voltage gain and input impedance. Full range for  $T_A$  is -55°C to 125°C for TLC372M, 0°C to 70°C for TLC372C.

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time	$R_L$ connected to 5 V through 5.1 k $\Omega$ , $C_L = 15\text{ pF}$ † See Note 6	100-mV input step with 5-mV overdrive	0.9		$\mu\text{s}$
		TTL-level input step	0.2		

† $C_L$  includes probe and jig capacitance.

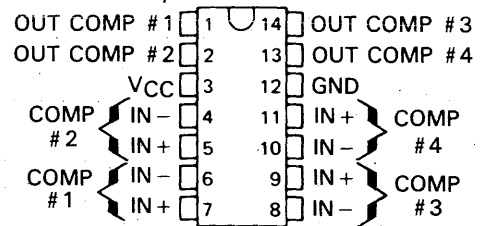
NOTE 6: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

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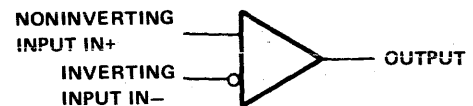
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- Single- or Dual-Supply Operation
- Wide Range of Supply Voltages  
2 to 18 volts
- Very Low Supply Current Drain  
0.4 mA Typ
- Fast Response Time . . . 200 ns Typ for  
TTL-Level Input Step
- Built-In ESD Protection
- High Input Impedance . . .  $10^{12} \Omega$  Typ
- Extremely Low Input Bias Current  
1 pA Typ
- Ultra-Stable Low Input Offset Voltage
- Common-Mode Input Voltage Range  
Includes Ground
- Output Compatible with TTL, MOS, and  
CMOS

TLC374M . . . J DUAL-IN-LINE PACKAGE  
TLC374C . . . D, J, OR N DUAL-IN-LINE PACKAGE  
(TOP VIEW)



symbol (each comparator)



**description**

This device is fabricated using LinCMOST™ technology and consists of four independent voltage comparators designed to operate from a single power supply. Operation from dual supplies is also possible so long as the difference between the two supplies is 2 to 18 volts. Each of these devices features extremely high input impedance (typically greater than  $10^{12}$  ohms) allowing direct interfacing with high-impedance sources. The outputs are n-channel open-drain configurations, and can be connected to achieve positive-logic wired-AND relationships. The TLC374C is designed as a pin-compatible, functional replacement for the LM339, offering twice the speed while consuming typically one-half of the power.

These devices have internal Electrostatic Discharge (ESD) protection circuits which will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883B Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

The TLC374M is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The TLC374C is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

**PRODUCT PREVIEW**

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# TYPES TLC374M, TLC374C QUADRUPLE LinCMOST™ DIFFERENTIAL COMPARATORS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TLC374M	TLC374C	UNIT
Supply voltage (see Note 1)	18	18	V
Differential input voltage (see Note 2)	±18	±18	V
Input voltage, $V_I$	18	18	V
Output voltage, $V_O$	18	18	V
Output current, $I_O$	20	20	mA
Duration of output short-circuit to ground (see Note 3)	unlimited	unlimited	
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 4)	900	900	mW
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J package		°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package		°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.  
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.  
 3. Short circuits from outputs to  $V_{CC}$  can cause excessive heating and eventual destruction.  
 4. For operation above 25°C free-air temperature, refer to Dissipation Derating Curves, Section 2. In the J package, the TLC374C chips are glass-mounted; the TLC374M chips are alloy-mounted.

electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC374M			TLC374C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = V_{ICR}$ min. See Note 5	25°C	2	10	2	10	mV	
		Full range		12		12		
$I_{IO}$ Input offset current	See Note 5	25°C	1		1		pA	
		Full range		10		0.3	nA	
$I_{IB}$ Input bias current		25°C	1		1		pA	
		Full range		20		0.6	nA	
$V_{ICR}$ Common-mode input voltage range		25°C	0 to $V_{CC} - 1.5$		0 to $V_{CC} - 1.5$		V	
		Full range	0 to $V_{CC} - 2$		0 to $V_{CC} - 2$			
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$ , $R_L \geq 15\text{ k}\Omega$ to $V_{CC}$	25°C	200		200		v/mV	
$I_{OH}$ High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C	0.1		0.1	nA	
		$V_{OH} = 15\text{ V}$	Full range		1		1	$\mu\text{A}$
$V_{OL}$ Low-level output voltage	$V_{ID} = -1\text{ V}$ , $I_{OL} = 4\text{ mA}$	25°C	150	400	150	400	mV	
		Full range		700		700		
$I_{OL}$ Low-level output current	$V_{ID} = -1\text{ V}$ , $V_{OL} = 1.5\text{ V}$	25°C	6	16	6	16	mA	
$I_{CC}$ Supply current (four comparators)	$V_{ID} = -1\text{ V}$ , No load	25°C	0.4	1	0.4	1	mA	

† All characteristics are measured with zero common-mode input voltage unless otherwise specified.

NOTE 5: The offset voltages and offset currents given are the maximum values required to drive the output up to 4 V or down to 400 mV with a pull-up resistor of 2.5 k $\Omega$  to  $V_{CC}$ . Thus, these parameters actually define an error band and take into account the worst-case effects of voltage gain and input impedance. Full range for  $T_A$  is -55°C to 125°C for TLC374M, 0°C to 70°C for TLC374C.

switching characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	$R_L$ connected to 5 V through 5.1 k $\Omega$ , $C_L = 15\text{ pF}$ †, See Note 6	100-mV input step with 5-mV overdrive		0.9		$\mu\text{s}$
		TTL-level input step		0.2		

†  $C_L$  includes probe and jig capacitance.

NOTE 6: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

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- Very Low Power Consumption . . . 1 mW  
Typ at  $V_{DD} = 5\text{ V}$
- Capable of Very High-Speed Operation  
. . . Typically 2 MHz in Astable Mode
- Complementary CMOS output Capable of  
Swinging Rail-to-Rail
- High Output-Current Capability  
. . . Sink 100 mA Typ  
. . . Source 10 mA Typ
- Output Fully CMOS-, TTL-, and  
MOS-Compatible
- Low Supply Current Reduces Spikes During  
Output Transitions
- High Impedance Inputs . . .  $10^{12}\ \Omega$  Typ
- Single-Supply Operation from 2 to 18 volts
- Functionally Interchangeable with the  
Signetics NE555; has Same Pinout

**description**

The TLC555 is a monolithic timing circuit fabricated using TI's LinCMOST™ process. Due to its high-impedance inputs (typically  $10^{12}\ \Omega$ ), it is capable of producing accurate time delays and oscillations while using less expensive, smaller timing capacitors than the NE555. Like the NE555, the TLC555 achieves both monostable (using one resistor and one capacitor) and astable (using two resistors and one capacitor) operation. In addition, 50% duty cycle astable operation is possible using only a single resistor and one capacitor. The LinCMOST™ process allows the TLC555 to operate at frequencies up to 2 MHz and be fully compatible with CMOS, TTL, and MOS logic. It also provides very low power consumption (typically 1 mW at  $V_{DD} = 5\text{ V}$ ) over a wide range of supply voltages ranging from 2 volts to 18 volts.

Like the NE555, the threshold and trigger levels are normally two-thirds and one-third respectively of  $V_{DD}$ . These levels can be altered by use of the control voltage terminal. When the trigger input falls below trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. When the reset input goes low, the flip-flop is reset and the output goes low. Whenever the output is low, a low impedance path is provided between the discharge terminal and ground.

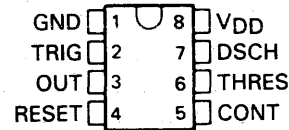
While the complementary CMOS output is capable of sinking over 100 mA and sourcing over 10 mA, the TLC555 exhibits greatly reduced supply current spikes during output transitions. This minimizes the need for the large decoupling capacitors required by the NE555.

These devices have internal Electrostatic Discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883B Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

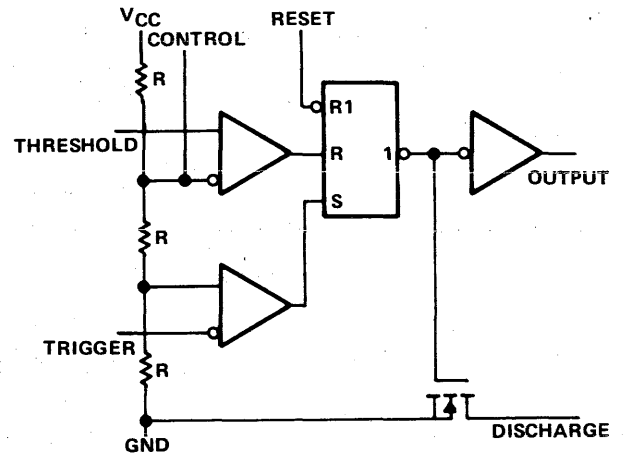
All unused inputs should be tied to an appropriate logic level to prevent false triggering.

The TLC555M is characterized for operation over the full military temperature range of  $-55\text{ }^\circ\text{C}$  to  $125\text{ }^\circ\text{C}$ ; the TLC555C is characterized for operation from  $0\text{ }^\circ\text{C}$  to  $70\text{ }^\circ\text{C}$ .

TLC555M . . . JG PACKAGE  
TLC555C . . . D, JG, or P PACKAGE  
(TOP VIEW)



**functional block diagram**



Reset can override Trigger, which can override Threshold.

**ADVANCE INFORMATION**

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# TYPES TLC555M, TLC555C LinCMOST™ TIMERS

FUNCTION TABLE

RESET	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
Low	Irrelevant	Irrelevant	Low	On
High	< 1/3 V <sub>DD</sub>	Irrelevant	High	Off
High	> 1/3 V <sub>DD</sub>	> 2/3 V <sub>DD</sub>	Low	On
High	> 1/3 V <sub>DD</sub>	< 2/3 V <sub>DD</sub>	As previously established	

† Voltages levels shown are nominal.

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, V <sub>DD</sub> (see Note 1)	18 V
Input voltage range (any input)	-0.3 V to 18 V
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2)	600 mW
Operating free-air temperature range: TLC555M	-55°C to 125°C
TLC555C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: D or P package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

2. For operation above 25°C free-air temperature, refer to Dissipation Derating Table. In the JG package, TLC555M chips are alloy-mounted and TLC555C chips are glass-mounted.

DISSIPATION DERATING TABLE

PACKAGE	POWER RATING	OPERATING FACTOR	ABOVE T <sub>A</sub>
D	600 mW	5.8 mW/°C	46°C
JG (Alloy-Mounted Chip)	600 mW	8.4 mW/°C	79°C
JG (Glass-Mounted Chip)	600 mW	6.6 mW/°C	59°C
P	600 mW	No derating necessary	

**electrical characteristics at 25°C free-air temperature, V<sub>DD</sub> = 5 V to 15 V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Threshold voltage level as a percentage of supply voltage			66.7%			
Threshold current	V <sub>DD</sub> = 5 V		10			pA
Trigger voltage level as a percentage of supply voltage			33.3%			
Trigger current	V <sub>DD</sub> = 5 V		10			pA
Reset voltage level			0.7			V
Reset current	V <sub>DD</sub> = 5 V		±10			pA
Control voltage (open-circuit) as a percentage of supply voltage			66.7%			
Low-level output voltage	V <sub>DD</sub> = 15 V	I <sub>OL</sub> = 10 mA	0.1			V
		I <sub>OL</sub> = 50 mA	0.5			
		I <sub>OL</sub> = 100 mA	1			
	V <sub>DD</sub> = 5 V	I <sub>OL</sub> = 5 mA	0.1			
		I <sub>OL</sub> = 8 mA	0.16			
High-level output voltage	V <sub>DD</sub> = 15 V	I <sub>OH</sub> = -1 mA	14.8			V
		I <sub>OH</sub> = -5 mA	14			
		I <sub>OH</sub> = -10 mA	12.7			
	V <sub>DD</sub> = 5 V	I <sub>OH</sub> = -2 mA	4			
		I <sub>OH</sub> = -1 mA	4.5			
Supply current	V <sub>DD</sub> = 15 V		360			μA
	V <sub>DD</sub> = 5 V		170			

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INSTRUMENTS

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operating characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Initial error of timing interval	$V_{DD} = 5\text{ V to }15\text{ V}$ , $R_A = R_B = 1\text{ k}\Omega\text{ to }100\text{ k}\Omega$ , $C_T = 0.1\ \mu\text{F}$ , See Figure 1		1%		
Supply voltage sensitivity of timing interval			0.1		%/V
Output pulse rise time	$V_{DD} = 5\text{ V}$ , $R_L = 10\text{ M}\Omega$ ,		20		ns
Output pulse fall time	$C_L = 10\text{ pF}$		20		
Maximum frequency in astable mode	$R_A = 470\ \Omega$ , $R_B = 200\ \Omega$ , $C_T = 200\text{ pF}$		2.1		MHz

TYPICAL APPLICATION DATA

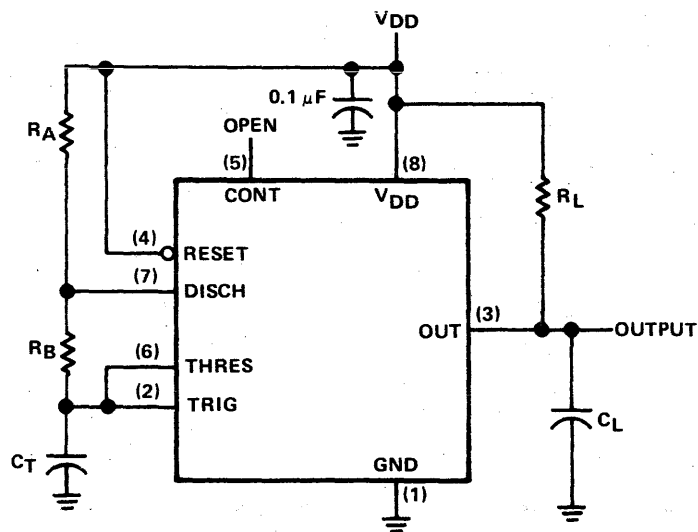


FIGURE 1—CIRCUIT FOR ASTABLE OPERATION

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# TDC1018

## Advance Information



### D/A Converter

8 bit, 125 MSPS

The TRW TDC1018 is a 125 MegaSample Per Second (MSPS), 8-bit digital-to-analog converter, capable of directly driving a 75-ohm load. Most applications require no extra registering, buffering, or deglitching. Four special level controls make the device ideal for video applications. All data and control inputs are ECL compatible.

The TDC1018 is built with TRW's OMICRON-B™ 1-micron dual-layer metal bipolar process. On-chip data registers and precise matching of propagation delays make the TDC1018 inherently low-glitching. The TDC1018 offers high performance, low power consumption, and video compatibility in a 24 Lead DIP package.

#### Features

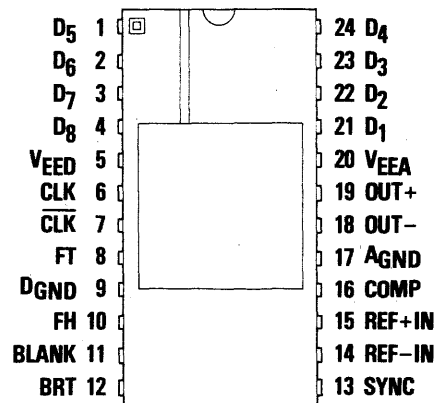
- 125 MSPS Conversion Rate
- 8-Bit
- 1/2 LSB Linearity
- Differential Current Outputs

- Video Controls: Sync, Blank, Bright, Force High
- Low Glitch Energy
- ECL Compatible
- Power Dissipation < 800 mW
- Available In 24 Lead DIP Package
- Single -5.2V Power Supply

#### Applications

- RGB Graphics
- Digital Synthesizers
- Automated Test Equipment
- Raster Graphic Displays
- High-Resolution Video
- Digital Transmitters/Modulators

#### Pin Assignments





# TDC1048

Preliminary Information



## Monolithic Video A/D Converter

8 bit, 20 MSPS

The TRW TDC1048 is a 20 MegaSample Per Second (MSPS) full-parallel (flash) analog-to-digital converter, capable of converting an analog signal with full-power frequency components up to 7 MHz into 8-bit digital words. A sample-and-hold circuit is not necessary. Low power consumption eases thermal considerations, and board space is minimized with a 28 Lead package. All digital inputs and outputs are TTL compatible.

The TDC1048 consists of 255 clocked latching comparators, combining logic, and an output buffer register. A single convert signal controls the conversion operation. The unit can be connected to give either true or inverted outputs in binary or offset two's complement coding.

### Features

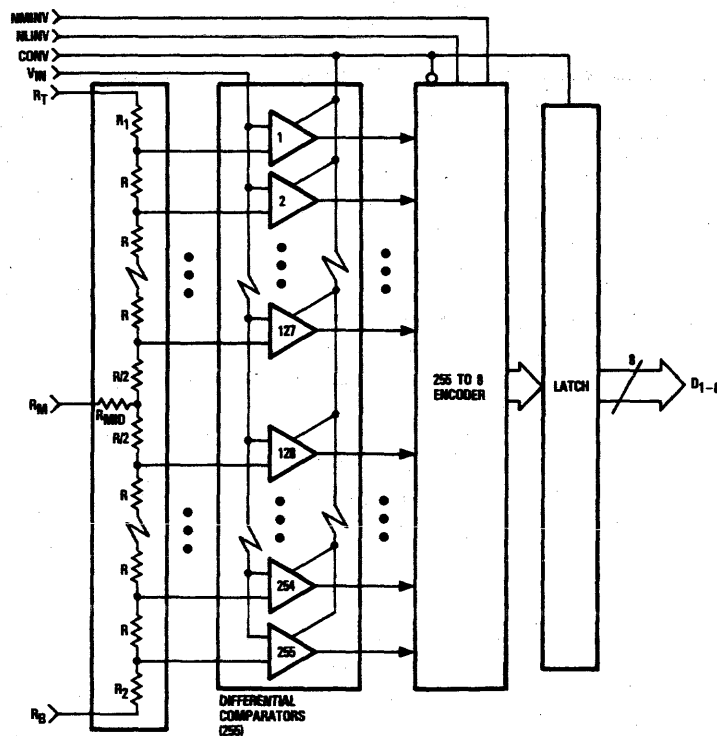
- 8-Bit Resolution
- 20 MSPS Conversion Rate

- Low Power Consumption, 1.4W (Worst Case)
- Sample-And-Hold Circuit, Not Required
- Differential Phase  $1^\circ$
- Differential Gain 2%
- 1/2 LSB Linearity
- TTL Compatible
- Selectable Output Format
- Available In 28 Lead DIP, CERDIP, Or Contact Chip Carrier

### Applications

- Low-Cost Video Digitizing
- Radar Data Conversion
- Data Acquisition
- Medical Imaging

### Functional Block Diagram



LSI Products Division  
TRW Electronic Components Group  
P.O. Box 2472  
La Jolla, CA 92038

Phone: (619) 457-1000  
Telex: 697-957  
TWX: 910-335-1571

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## A/D Converters

All TRW A/D converters, except the TDC1001 and TDC1002, are fully parallel "flash" types, which accurately sample and convert high-frequency input signals without a track-and-hold circuit. Resolutions from 4 to 9 bits and speeds from 1 to 100 MSPS (MegaSamples Per Second) offer the best choice for most signal processing requirements. The parts are TTL compatible, except as noted below.

Product	Resolution (bits)	Conversion Rate* (MSPS)	Power Dissipation* (watts)	Package	Available	Extended Temperature Range Available	Notes
TDC1001	8	2.5	0.6	J8	Now	-20°C to 85°C	Successive approximation
TDC1002	8	1.0	0.6	J8	Now	-20°C to 85°C	Successive approximation
TDC1007	8	20	2.6	J1, C1, L1 E1, P1	Now Now	-30°C to 125°C no	Evaluation boards
TDC1014	6	25	1.0	J7, C3, B7 E1, P1	Now Now	-55°C to 125°C no	Evaluation boards
TDC1019	9	15	2.6	J1, C1, L1 E1	Now Now	3Q83 no	ECL-compatible Evaluation board
TDC1021	4	25	0.3	J9	Now	-55°C to 125°C	
TDC1025	8	50	4.5	C1, L1 E1	3Q83 Now	3Q83 no	ECL-compatible Evaluation board
TDC1027	7	18	1.6	J7, B7	Now	4Q83	
TDC1029	6	100	1.6	J7 E1	Now Now	3Q83 no	ECL-compatible Evaluation board

\*Guaranteed, Worst Case,  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ .

## D/A Converters

These D/A converters offer 75-ohm, 1V outputs to drive video filters or transmission lines directly. On-chip data registers and careful internal timing to matched current sources result in output glitches that are so small (less than 100 picovolt-seconds worst case) that no resampling circuit or deglitcher is necessary. The TDC1016 has 10 input data bits and is available in three integral linearity grades equivalent to 1/2 LSB error at 8, 9, and 10 bit resolution. The parts have dual-level input buffers which operate with TTL or ECL signals: In ECL mode, only a single -5V power supply is required. The TDC1016 in the J5 or C2 package can operate in a fully differential ECL mode, as well as single-ended ECL or TTL.

Product	Bits	Integral Linearity Error (%)	Conversion Rate * (MSPS)	Power Dissipation* (watts)	Package	Available	Extended Temperature Range Available	Notes
TDC1016	8**	0.20	20	0.7	J5, J7, C2, B7	Now	-55°C to 125°C	
	9**	0.10	20	0.7	J5, J7, C2, B7	Now	3Q83	
	10**	0.05	20	0.7	J5, J7, C2, B7	Now	3Q83	

\* Guaranteed, Worst Case,  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ .

\*\*The TDC1016 has 10 bit resolution, and is available in three linearity grades to meet 8, 9, and 10 bit system requirements.

# TDC1025

## Preliminary Information



### Monolithic A/D Converter

8 Bit, 50 MSPS

The TRW TDC1025 is a 50 MegaSample Per Second (MSPS) full-parallel (flash) analog-to-digital converter, capable of converting an analog signal with full-power frequency components up to 12 MHz into 8-bit digital words. A sample-and-hold circuit is not necessary. All digital inputs and outputs are ECL compatible.

The TDC1025 consists of 255 latching comparators, combining logic, and an output buffer register. A single convert signal controls the conversion operation. The digital outputs will interface with differential or single-ended ECL. The device requires a single  $-5.2V$  power supply.

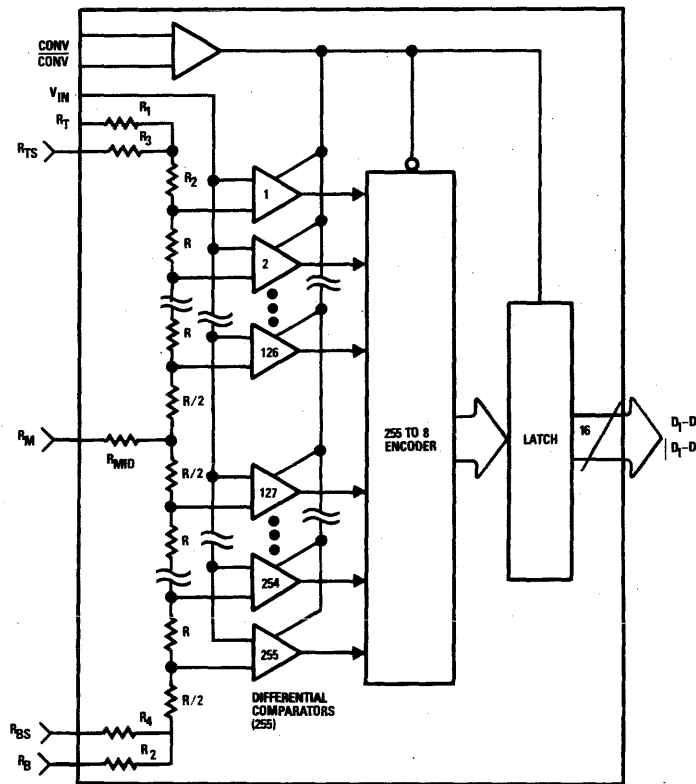
### Features

- 8-Bit Resolution
- 50 MSPS Conversion Rate
- Sample-and-Hold Circuit Not Required
- Differential or Single-Ended ECL Compatible
- Available in 68 Leaded Chip Carrier and 68-Contact Chip Carrier
- Advanced 1-micron (OMICRON-B™) Technology
- Complete Evaluation Board (TDC1025E1C) Available

### Applications

- Medical Electronics
- Fluid Flow Analysis
- Seismic Analysis
- Radar/Sonar
- Transient Analysis
- High-Speed Image Processing

### Functional Block Diagram



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LSI Products Division  
 TRW Electronic Components Group  
 P.O. Box 2472  
 La Jolla, Ca. 92038

Phone: (619) 457-1000  
 Telex: 697-957  
 TWX: 910-335-1571

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### Monolithic Video A/D Converter

7 Bit, 18 MSPS

The TRW TDC1027 is an 18 MegaSample Per Second (MSPS) full-parallel (flash) analog-to-digital converter, capable of converting an analog signal with full-power frequency components up to 5 MHz into 7-bit digital words. A sample-and-hold circuit is not necessary. All digital inputs and outputs are TTL compatible.

The TDC1027 consists of 127 latching comparators, combining logic, and an output buffer register. A single convert signal controls the conversion operation. The unit can be connected to give either true or inverted outputs in binary or offset two's complement coding.

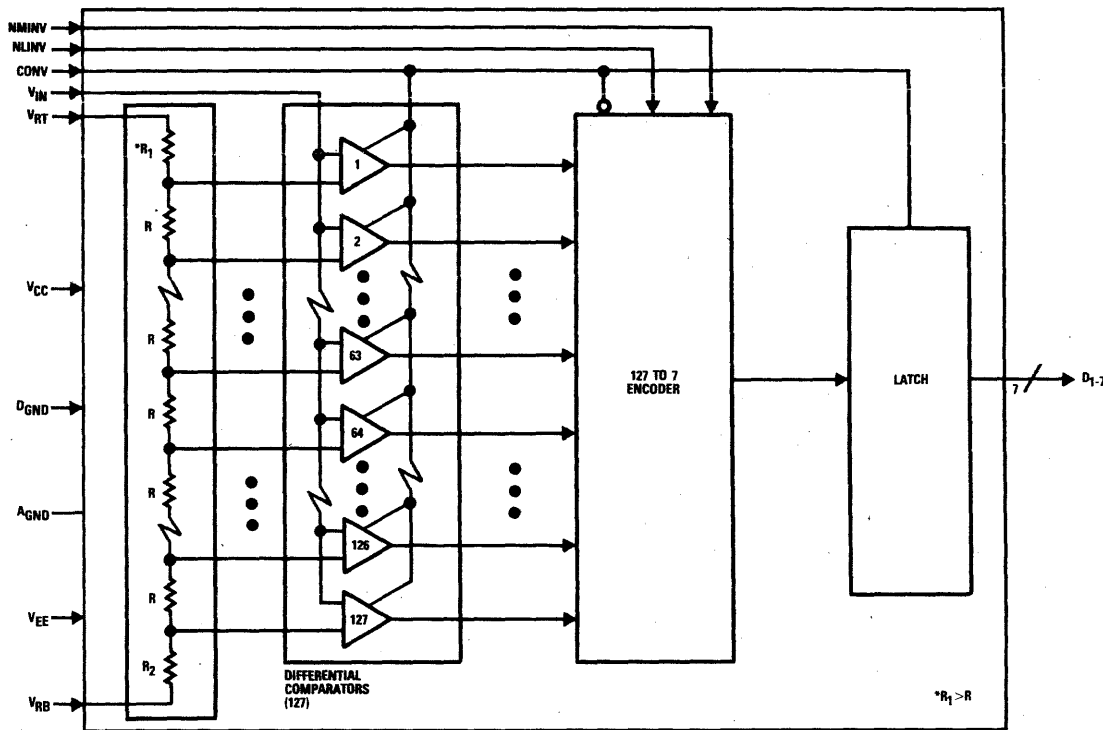
### Features

- 7-Bit Resolution
- 1/2 LSB Linearity
- Sample-And-Hold Circuit Not Required
- TTL Compatible
- 18 MSPS Conversion Rate
- Selectable Output Format
- Available in 24 Lead DIP

### Applications

- Low-Cost Video Digitizing
- Medical Imaging
- Data Acquisition
- TV Special Effects
- Video Simulators
- Radar Data Conversion

### Functional Block Diagram



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**LSI Products Division**  
 TRW Electronic Components Group  
 P.O. Box 2472  
 La Jolla, Ca. 92038

Phone: (619) 457-1000  
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# TDC1029

## Preliminary Information



### Monolithic A/D Converter

6 bit, 100 MSPS

The TRW TDC1029 is a 100 MegaSample Per Second (MSPS) full-parallel (flash) analog-to-digital converter, capable of converting an analog signal with full power frequency components up to 50 MHz into 6-bit digital words. A sample-and-hold circuit is not necessary. All digital inputs and outputs are ECL compatible.

The TDC1029 consists of 63 clocked latching comparators, combining logic, and an output buffer register. A single differential convert signal controls the conversion operation. The digital outputs are single-ended ECL with the exception of the MSB which is differential. Binary or offset two's complement output format is available.

#### Features

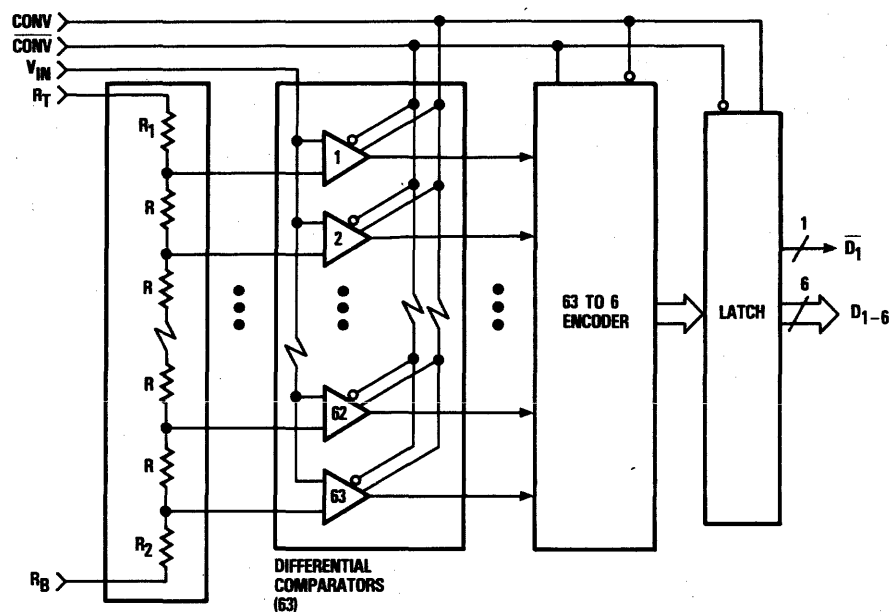
- 6-Bit Resolution
- 100 MSPS Conversion Rate
- 50 MHz Input Bandwidth

- Low Cost
- 1/2 LSB Linearity
- Sample-And-Hold Circuit: Not Required
- 1V Input Range
- Aperture Error < 50 psec
- Binary Or Two's Complement Output Format
- Available In 24 Lead DIP

#### Applications

- Transient Digitizers
- Direct Digital Receivers
- Radar Data Conversion
- Data Acquisition
- Telecommunications
- Medical Imaging

#### Functional Block Diagram



LSI Products Division  
 TRW Electronic Components Group  
 P.O. Box 2472  
 La Jolla, CA 92038

Phone: (619) 457-1000  
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# LINEAR INTEGRATED CIRCUITS

## Pulse Width Modulators



TYPE	PERFORMANCE CHARACTERISTICS																	
	Voltage Reference $\pm 4\%$	Voltage Reference $\pm 1\%$	Soft Start	PWM Latch	Under-Voltage Lockout	Pulse-by-Pulse Current Limiting	Shutdown Terminal	Output Current	Feed Forward	Maximum Frequency Oscillator	Dual Uncommitted Outputs	Single-Ended Outputs	Totem Pole Outputs	Separate Oscillators	Adjustable Oscillator Sync Terminal	Latch Off or Continuous Control	Double Pulse Suppression	Low Current Start Up
<b>Regulating PWMs</b> UC1524 UC1524/883B	X							100mA		300kHz	X							
<b>Advanced Regulating PWMs</b> UC1524A UC1524A/883B	X		X	X	X	X		200mA		500kHz	X							X
<b>Advanced Regulating PWMs</b> UC1525A UC1525A/883B	X	X	X	X		X		100mA 0.4A Pulse		500kHz		X	X	X				
<b>Advanced Regulating PWMs</b> UC1527A UC1527A/883B	X	X	X	X		X		100mA 0.4A Pulse		500kHz		X	X	X				
<b>Advanced Regulating PWMs</b> UC1526 UC1526/883B	X	X	X	X	X	X		100mA		400kHz		X	X	X				X
<b>Regulating PWMs</b> UC493/UC494/UC495	X							200mA		300kHz	X			X				X
<b>Advanced Regulating PWMs</b> UC493A/UC494A/UC495A* UC493AC/UC494AC/UC495AC*	X					X		200mA		300kHz	X			X				X
<b>Current Mode PWM Controllers</b> UC1846 UC1846/883B	X	X	X	X	X	X		200mA	X	500kHz		X	X	X	X			X
<b>Current Mode PWM Controllers</b> UC1847 UC1847/883B	X	X	X	X	X	X		200mA	X	500kHz		X	X	X	X			X
<b>Programmable Primary Side PWMs</b> UC1840 UC1840/883B	X	X	X	X	X	X		200mA	X	500kHz	X			X	X	N/A		X
<b>Programmable Primary Side PWMs</b> UC1842 UC1842/883B	X		X	X	X			100mA 1A Pulse	X	500kHz	X			X		N/A		X

\*Series available with 883B screening.

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TYPE	DESCRIPTION	KEY FEATURES	PACKAGE
UC1543/2543/3543 UC1544/2544/3544	Power Supply Supervisory Circuit, Monitors and Controls Power Supply Output	<ul style="list-style-type: none"> <li>• Over/Under-Voltage, and Current Sensing Circuits</li> <li>• Programmable Time Delays</li> <li>• SCR "Crowbar" Drive of 300mA</li> <li>• Optional Over-Voltage Latch</li> <li>• Internal 1% Accurate Reference</li> <li>• Remote Activation Capability</li> <li>• Uncommitted Comparator</li> <li>• Inputs for Low Voltage Sensing (UC1544 series only)</li> </ul>	16 Pin DIL (1543 Series)  18 PIN DIL (1544 series)
UC1706/2706/3706	Dual High Current MOSFET Compatible Output Driver	<ul style="list-style-type: none"> <li>• Dual, 1.5A, Totem Pole Outputs</li> <li>• Parallel or Push-Pull Operations</li> <li>• Single-Ended to Push-Pull Conversion</li> <li>• Internal Overlap Protection</li> <li>• Analog, Latched Shutdown</li> <li>• High-Speed, Power MOSFET Compatible</li> <li>• Thermal Shutdown Protection</li> <li>• 5 to 40V Operation</li> <li>• Low Quiescent Current</li> </ul>	16 Pin DIL
UC1834/2834/3834	High Efficiency Linear Regulator, Low Input-Output Differential	<ul style="list-style-type: none"> <li>• Minimum <math>V_{IN}-V_{OUT}</math> less than 0.5V at 5A Load with External Pass Device</li> <li>• Equally Usable for either Positive or Negative Regulator Design</li> <li>• Adjustable Low Threshold Current Sense Amplifier</li> <li>• Under- and Over-Voltage Fault Alert with Programmable Delay</li> <li>• Over-Voltage Fault Latch with 100mA Crowbar Drive Output</li> </ul>	16 Pin DIL
UC1901/2901/3901	Isolated Feedback Generator Stable and Reliable Alternative to an Optical Coupler	<ul style="list-style-type: none"> <li>• An Amplitude-Modulation System for Transformer Coupling an Isolated Feedback Error Signal</li> <li>• Internal 1% Reference and Error Amplifier</li> <li>• Loop Status Monitor</li> <li>• Low-Cost Alternative to Optical Couplers</li> <li>• Internal Carrier Oscillator Usable to 5MHz</li> <li>• Modulator Synchronizable to an External Clock</li> </ul>	14 Pin DIL
UC1903/2903/3903	Quad Supply and Line Monitor Precision System	<ul style="list-style-type: none"> <li>• Monitor Four Power Supply Output Voltage Levels</li> <li>• Both Over- and Under-Voltage Indicators</li> <li>• Internal Inverter for Negative Level Sense</li> <li>• Adjustable Fault Window</li> <li>• Additional Input for Early Line Fault Sense</li> <li>• On Chip, High-Current General Purpose OP-AMP</li> </ul>	18 Pin DIL

**Motor Control Circuits**

TYPE	DESCRIPTION	KEY FEATURES	PACKAGE
UC1637/2637/3637	Switched Mode Controller for DC Motor Drive	<ul style="list-style-type: none"> <li>• Single or Dual Supply Operation</li> <li>• <math>\pm 2.5V</math> to <math>\pm 20V</math> Input Supply Range</li> <li>• <math>\pm 5\%</math> Initial Oscillator Accuracy; <math>\pm 10\%</math> Over Temperature</li> <li>• Pulse-by-Pulse Current Limiting</li> <li>• Under-Voltage Lockout</li> <li>• Uncommitted PWM Comparators for Design Flexibility</li> </ul>	18 Pin DIL
UC3717	Stepper Motor Drive Circuit	<ul style="list-style-type: none"> <li>• Half-Step and Full-Step Mode</li> <li>• Bipolar Constant Current Motor Drive</li> <li>• Built-In Protection Diodes</li> <li>• Wide Range of Current Control 5-1000mA</li> <li>• Wide Voltage Range 10-45V</li> <li>• Designed for Unregulated Motor Supply Voltage</li> <li>• Thermal Overload Protection</li> </ul>	16 Pin DIL



## Three Terminal Voltage Regulators, Adjustable

TYPE	OUTPUT CURRENT (A)	POLARITY	REGULATED OUTPUT VOLTAGE (V)	PACKAGE
UC117K/LM117K UC217K/LM217K *UC317K/LM317K	1.5A	Pos.	Adjustable from 1.2V to 37V	TO-3 TO-3 TO-3
UC137K/LM137K UC237K/LM237K *UC337K/LM337K	1.5A	Neg.	Adjustable from -1.2V to -37V	TO-3 TO-3 TO-3
UC150K/LM150K UC250K/LM250K *UC350K/LM350K	3.0A	Pos.	Adjustable from 1.2V to 33V	TO-3 TO-3 TO-3

## Three Terminal Voltage Regulators, Fixed, Positive

TYPE	OUTPUT CURRENT (A)	POLARITY	REGULATED OUTPUT VOLTAGE (V)			PACKAGE
UC7800AK/LM140AK SERIES *UC7800ACK/LM340AK SERIES	1.5A	Pos.	5V ± 1%	12V ± 1%	15V ± 1%	TO-3 TO-3
UC7800K/LM140K SERIES *UC7800CK/LM340K SERIES	1.5A	Pos.	5V ± 4%	12V ± 4%	15V ± 4%	TO-3 TO-3

## Three Terminal Voltage Regulators, Fixed, Negative

TYPE	OUTPUT CURRENT (A)	POLARITY	REGULATED OUTPUT VOLTAGE (V)			PACKAGE
UC7900AK/LM120K SERIES *UC7900ACK SERIES	1.5A	Neg.	-5V ± 1%	-12V ± 1%	-15V ± 1%	TO-3 TO-3
UC7900K SERIES *UC7900CK/LM320K SERIES	1.5A	Neg.	-5V ± 4%	-12V ± 4%	-15V ± 4%	TO-3 TO-3

\* Also available in TO-220 package.

# INTRODUCTION TO MEMORY

The Master Selection Guide provides sufficient information for making initial product selections. All devices that appear in this section, both in the initial selection guide and the data pages, are included in the indexes. These index listings lead to the page and line on that page where each device appears.

The Memory Section provides initial selection information and data on PROMs, RAMs, and ROMs as well as other memory devices. In these particular sections, the devices are characterized by organization (words and bits/word) and by access times. In order to assure that the access times are comparable, whenever possible the values have been shown in nanoseconds over the full rated temperature range for the devices (i.e., 0° to 70°C for commercial units and -55°C to 125°C for military units). The full temperature nanosecond value is marked "nsF." When this value isn't specified, the guaranteed nanosecond value at room temperature is listed followed by "nsR." In some cases a guaranteed value has not been established; then the typical value is shown followed by "ns\*": "Typical" values are invariably much faster than the guaranteed ones so that such listings place these memories higher on the list than they otherwise would appear.

CATEGORY	
Character Generators	3755
Code Converters	3756
FIFOs, LIFOs	3757
E <sup>2</sup> PROMS	3758
PROMs	3760
RAMs	
Bubble	3768
CCD	3768
Dynamic	3768
Static	3770
ROMs	3783
Shift Registers	
Dynamic	3788
Static	3788

## Detailed Product Information provided by:

Advanced Micro Devices	3901
American Microsystems, Inc.	3932
Fujitsu America	3938
Harris Semiconductor	3943
Hitachi America Ltd.	3970
Inmos	3977
Integrated Device Technology	3982
Intel	4021
Monolithic Memories, Inc.	4026
Motorola Semiconductor	4046
National Semiconductor	4057
NEC Electronics	4073
OKI Semiconductor	4118
Panasonic	4120
Raytheon	4130
SEEQ Technology	4133
Signetics	4141
Silicon Systems	4208
Synertek	4210
Texas Instruments	4212
VLSI Technology	4249
Western Digital	4260
Xicor	4265

The manufacturers listed above have provided detailed information on their latest and most significant products.

# EINFÜHRUNG SPEICHER

Der Master Selection Guide für Speicher enthält alle Informationen, die Sie für die Erstausswahl Ihres Produkts benötigen. Die Bauteile, die in diesem Abschnitt erscheinen, sowohl im Selection Guide als auch auf den Datenblättern, sind in allen Master Indexes enthalten. Diese Register verweisen auf die Seite und Zeile, auf der das entsprechende Bauelement vorkommt.

Der Speicherteil besteht aus einem Selection Guide und Daten über PROMs, RAMs und ROMs sowie andere Speicher-Bauteile. In diesen einzelnen Abschnitten sind die Bauteile nach Organisation (Worte und Bits/Wort) und nach Zugriffszeiten eingeteilt. Um sicherzustellen, daß die Zugriffszeiten vergleichbar sind, wurden die Werte nach Möglichkeit in Nanosekunden für den vollen zugelassenen Temperaturbereich (d.h. 0° bis 70°C für kommerzielle Bauteile und -55° bis 125°C für militärische Bauteile) angegeben. Der nsec-Wert für den vollen Temperaturbereich ist mit "nsF" bezeichnet. Wenn dieser Wert nicht genannt ist, so ist der garantierte nsec-Wert bei Raumtemperatur mit "nsR" angegeben. In einigen Fällen wurde ein garantierter Wert nicht festgelegt. Dann wird der typische Wert mit einem nachfolgenden "ns\*" markiert. "Typische" Werte sind durchwegs schneller als garantierte, so daß solche Eintragungen die Speicher weiter oben in der Liste erscheinen lassen als dies sonst der Fall wäre.

# INTRODUCTION AUX MEMOIRES

Le Guide Général de Sélection fournit suffisamment de renseignements pour permettre des sélections initiales de produits. Tous les appareils énumérés dans cette section, tant dans le Premier Guide de Sélection que dans les feuilles de données, sont inclus dans tous les index. Ces index vous procurent la possibilité de retrouver à quelle page et à quelle ligne tel ou tel appareil a été mentionné.

La Section "Mémoires" contient des données et des renseignements de sélection initiale sur les PROMs, les RAMs et les ROMs, ainsi que sur d'autres appareils mémoires. Dans ces sous-sections, les appareils sont distingués par définition (mots et bit/mot) et par temps d'accès. Afin d'être sûr que les temps d'accès soient comparables, chaque fois que cela a été possible, les valeurs ont été indiquées en nanosecondes pour toute la fourchette de température (exemple: 0° à 70° pour les produits industriels, et -55°C à 125°C pour les produits militaires). La valeur entière de température en nanosecondes est désignée par "nsf". Lorsque cette valeur n'est pas donnée, une valeur garantie à température ambiante, exprimée en nanosecondes, est suivie par "nsR". Dans certains cas aucune valeur garantie en nanosecondes n'a pu être établie, la valeur "typique" est alors traduite par "ns". Les valeurs "typiques" sont invariablement beaucoup plus rapides que celles dites "garanties", de sorte que ces mémoires "typiques" apparaissent plus haut sur la liste qu'elles ne le seraient sur d'autres.

# INTRODUCCIÓN A MEMORIA

La Guía Maestra de Selección provee suficiente información para hacer selecciones iniciales del producto. Todas las componentes que aparecen en esta sección, ya sea en la guía de selección inicial o en las páginas de datos, están incluidas en todos los otros índices. Estas listas de índices los conduce a la página y línea de aquella página donde se encuentra cada componente.

La Sección de Memoria provee información para la selección inicial y datos de PROMs, RAMs, and ROMs así como otras piezas de memoria. En estas secciones en particular, las componentes se caracterizan por organización (palabras y bits/palabras) y por tiempo de acceso. Con objeto de asegurar que los tiempos de acceso son comparables, cuando es posible, los valores aparecen en nanosegundos sobre el intervalo completo de temperatura para las componentes (en otras palabras, 0° to 70°C para piezas de uso comercial y -55° a 125°C para piezas de uso militar). El valor del intervalo completo de temperatura esta marcado "nsF". Cuando este valor no está especificado, el valor de nanosegundon está garantizado para temperatura ambiental y aparece seguido por "nsR". En algunos casos el valor no ha sido establecido; entonces el valor tipico aparece seguido por "ns\*". Valores "típicos" son sin falta más rápido que los garantizados de tal manera que las listas ubica estas memorias antes que normalmente los mostraría.

# メモリーへの案内

マスターセレクションガイドは選択に取りかかる十分な資料を揃えています。セレクションガイド、データ掲載ページにのっている製品は全てのインデックスに入っており、そのインデックスによりページ数、行数はすぐ見つけられます。

このセクションには PROM, RAM, ROM 他のメモリーのデータが入っています。製品はその構成(ワード, ビット/ワード)とアクセスタイム毎に分類されています。アクセスタイムが同等であるかどうか確認する為可能な限り夫々の全温度範囲(一般0℃~70℃; ミリタリ-55℃~125℃)でのアクセスタイムをナノセカンドで表示しています。全温度範囲でのアクセスタイムは "ns F" と表示, 規定されていないものすなわち室温での保証値は "ns R" となっています。保証値が決められていない場合は代表的値として "ns★" がついています。"Typical" 値は通常保証値よりも速くつけられるのでリスト上では他に記載されているものより高い場所に記載されています。

MEMORY-Character Generators

Format	Number of Output Lines	Input Logic levels	Supply Voltage, V	Device	Source	Line
<b>Character Generators</b>						
5x7	1	TTL	5	DM76S128 DM76S64 DM8678 DM86S128 DM86S64	† National † National National National National	5
	5	MOS TTL	5 5	MM52116FDW R03-2513 5055 5155 6055 6155 NCM6670 NCM6674	National GI MMI MMI MMI MMI Motorola (4048) Motorola (4048)	10
			-12,5	3258 32581 32582	Fairchild Fairchild Fairchild	15
	7	TTL	5	6056 6061 6062 6156 6161 6162 3257	MMI MMI MMI MMI MMI MMI Fairchild	20
	35	TTL	5	MSL9650	OKI (4118)	25
6x10	7	MOS	5	MCH-01	Motorola	25
7x9	1	TTL	5	DM76S128 DM76S64 DM8678 DM86S128 DM86S64	† National † National National National National	30
	7	MOS TTL	5 5	MM52116FDX 5290 5291 6071 6072 6171 6172 6290 6291 NCM66700 NCM66710 NCM66714 NCM66720 NCM66730 NCM66734 NCM66740 NCM66750 NCM66751 NCM66760 NCM66770 NCM66780 NCM66790	National † MMI † MMI MMI MMI MMI MMI MMI MMI Motorola (4048) Motorola (4048) Motorola (4048) Motorola (4048) Motorola (4048) Motorola (4048) Motorola (4048) Motorola (4048) Motorola Motorola (4048) Motorola (4048) Motorola (4048) Motorola (4048)	35 40 45 50
			-12,5	3260	Fairchild	55
7x11	1	TTL	5	CRT7004 CRT8002	SMC SMC	55
9x9	9	TTL	5	5292 5293 6292 6293	† MMI † MMI MMI MMI	60
10x9	10	TTL	5	SCR2670	Sigmetics (1677)	60
14 Segments	14	TTL	5	MSL9664 MSL9665	OKI (4118) OKI (4118)	
16 Segments	16	TTL	5	MSL9662 MSL9663	OKI (4118) OKI (4118)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MEMORY-Code Converters

Code Conversion From To	Bits In	Bits Out	Process	Supply Voltage, V	Device	Source	Line
<b>Code Converters</b>							
BINARY-BCD	6	6	Bipolar	5	<b>SN54185A</b> † TI (988) <b>SN74185A</b> TI (988)		
Degree-Sine					<b>5086</b> † MMI (819) <b>5087</b> † MMI (819) <b>6086</b> MMI (819) <b>6087</b> MMI (819)		5
EBCDIC-ASCII	12	8	Bipolar	5	<b>DM8576AAA</b> National		
Multiple ASCII-SELECTRIC, EBCDIC, HOLLERITH			NMOS	±5	<b>MCM68316</b> Motorola <b>MCM68A310E-01</b> Motorola (4048)		
Binary-BCD	9	7	Bipolar	5	<b>SN548485</b> † TI (1043) <b>SN748485</b> TI (1043)		10
BCD-BINARY	6	6	Bipolar	5	<b>SN54184</b> † TI (987) <b>SN74184</b> TI (987)		
BCD-Binary	7	9	Bipolar	5	<b>SN548484</b> † TI (1043) <b>SN748484</b> TI (1043)		15

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



MEMORY-FIFOs, LIFOs

Words	Bits/ Word	Data Rate MHz (max)	Supply Voltage, V	Device	Source	Line	Words	Bits/ Word	Data Rate MHz (max)	Supply Voltage, V	Device	Source	Line		
<b>First-in First-out</b>							<b>Last-in First-out</b>								
16	4	8.0 *	3-15	CD40105B	† RCA (838)	5	132/128	9	0.65	5	WD1510-00	Western	65		
				CD40105BE	RCA (838)		1.0		5		WD1510-01	Western			
			10.0	5	54F403		† Fairchild (743)	512	9	4.2	5	MK4501-20		Mostek	
					74F403		Fairchild (743)	5.7		5	MK4501-15	Mostek			
					W9403		Signetics (1660)	7.1		5	MK4501-12	Mostek			
			12.0	5	SN54LS222		† TI (995)	<b>Last-in First-out</b>							
					SN54LS224		† TI (995)	81	4	1.0	5	SR5018		SMC	70
					SN74LS222		TI (995)	132/128				9		1.0	
					SN74LS224		TI (995)	133	4	1.0	5				SR5017
			20.0	5	SN54LS227		† TI (997)								
				SN54LS228	† TI (997)										
				SN74LS227	TI (997)										
				SN74LS228	TI (997)										
	5	10.0	5	SN74S225	TI (996)										
24	4	2.5	5-16	MS618	† RTC	15									
32	8	0.5	-12.5	AM2812C	AMD	15									
				AM2812L	AMD										
				MJ2812	Plessey										
				MJ2814	† Plessey										
			1.0	-12.5	AM2812AC		AMD	20							
					AM2812AL		AMD								
		9	0.5	-12.5	AM2813C		AMD								
					AM2813L		AMD								
					MJ2813		Plessey								
			1.0	-12.5	AM2813AC		AMD	25							
				AM2813AL	AMD										
40	9	0.5	-12.5	FR1502-11	Western	30									
				33512	Fairchild										
				3351M	† Fairchild										
	FR1502-10	Western													
			1.5	-12.5	3351-3		Fairchild								
			2.0	-12.5	33511		Fairchild								
64	4	0.7	-12.5	AM3341C	AMD	35									
				3341	Fairchild										
				3341M	† Fairchild										
				AM2841C	AMD										
		AM2841M	† AMD												
		3341A	Fairchild												
		MJ2841	Plessey												
		AM2841A	AMD	40											
			1.2	-12.5	AM2841A	AMD									
			7.0	5	C57401	† MMI (805)	50								
			8.0	5	9423C	Fairchild									
					9423M	† Fairchild									
			10	5	DM77S401	† National	45								
					DM87S401	National									
			10.0	5	54F413	† Fairchild (750)									
					54F433	† Fairchild (743)									
				74F413	Fairchild (750)										
				74F433	Fairchild (743)										
				C57401A	† MMI (805)	50									
				C67401	MMI (805)										
		15	5	DM77S401A	† National										
				DM87S401A	National										
		15.0	5	C67401A	MMI (805)	55									
		16.7	5	C67401B	MMI (805)										
	5	7.0	5	C57402	† MMI (805)	60									
		10	5	DM77S402	† National										
				DM87S402	National										
		10.0	5	C57402A	† MMI (805)										
				C67402	MMI (805)										
		15	5	DM77S402A	National										
				DM87S402A	National										
		16.7	5	C67402B	MMI (805)										
	9	15.0	5	TDC1030	TRW-LSI (1161)										

† Military Temperature Range (-55° to 125°C)  
OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range  
TS—Three-State

nsR—Nanoseconds at Room Temperature  
OE—Open Emitter



MEMORY-E<sup>2</sup> PROMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
<b>E<sup>2</sup> PROMs</b>						<b>(Cont'd)</b>		
8192x4	450 nsF	E <sup>2</sup>	TS	28	5	NCR52864	NCR	
8192x8	200 nsF	E <sup>2</sup>	TS	28	5	XLS2864A-200	EXEL	
						52B33-200	SEEQ	(4133)
						52B33H-200	SEEQ	(4133)
						TMM2764	Toshiba	
	250 nsF	E <sup>2</sup>	TS	28	5	XLS2864A-250	EXEL	5
						M2764-25	Intel	
						52B33-250	SEEQ	(4133)
						52B33H-250	SEEQ	(4133)
	300 nsF	E <sup>2</sup>	TS	28	5	XLS2864A-300	EXEL	10
						52B33-3	SEEQ	(4133)
						52B33-300	SEEQ	(4133)
						52B33H-3	SEEQ	(4133)
						52B33H-300	SEEQ	(4133)
	350 nsF	E <sup>2</sup>	TS	28	5	XLS2864A-350	EXEL	15
						X2864A	Xicor	(4269)
	450 nsF	E <sup>2</sup>	TS	28	5	M2764-45	Intel	
						NMH2864	† National	
						NMH2864C	National	

† Military Temperature Range (-55° to 125°C)  
OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range  
TS—Three-State

nsR—Nanoseconds at Room Temperature  
OE—Open Emitter

MEMORY-PROMs

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
<b>PROMs</b>									256x4	10 nsF	ECL	OE	16	-5.2	F10Z416A	Fairchild	
1Kx8	55 nsF	NMOS	TS	24	5	MB7132E-W	† Fujitsu		11 ns *	ECL	OE	16	-5.2	F10416C	Fairchild		
2Kx4	35 nsF	NMOS	TS	20	5	MB7128Y	Fujitsu		12 ns *	ECL	OE	16	-5.2	GXB1049	Siemens		
	55 nsF	CMOS	TS	20	5	MB7128E-W	† Fujitsu		15 nsF	ECL	OE	16	-5.2	F100Z416	Fairchild		
2Kx8	55 nsF	NMOS	TS	24	5	MB7138E-W	† Fujitsu							F10Z416	Fairchild		65
4Kx4	35 nsF	NMOS	TS	20	5	MB7152-Y	Fujitsu	5	20 nsR	ECL	OE	16	-5.2	10149	Signetics (858)		
	55 nsF	NMOS	TS	20	5	MB7152E-W	† Fujitsu		24 nsF	ECL	OE	16	-4.5	F100416C	Fairchild		
8Kx8	70 nsF	NMOS	TS	24	5	MB7144E-W	† Fujitsu		25 nsF	ECL	OE	16	-5.2	MCM10149	Motorola (4047)		
16x4	25 nsF	TTL	TS	16	5	DM85S47A	National		30 nsF	TTL	OC	16	5	AM27S20AC	AMD		70
	30 nsF	TTL	TS	16	5	DM75S07A	† National							AM27S21AC	AMD		
	35 nsF	TTL	OC	16	5	DM85S06	National	10	35 nsF	TTL	OC	16	5	HM7610B-5	Harris (3944)		
	50 nsF	TTL	OC	16	5	DM75S07	† National							HM7611B-5	Harris (3944)		
32x8	15 nsF	TTL	OC	16	5	DM87S188	National		40 nsF	TTL	OC	16	5	AM27S20AM	† AMD		
	20 nsF	TTL	OC	16	5	DM77S188	† National							HM7620B-5	Harris (3945)		75
	20 nsR	ECL	OE	16	-5.2	SN10139	TI							AM27S21AM	† AMD		
	25 nsF	ECL	OE	16	-5.2	MCM10139	Motorola (4047)							HM7621B-5	Harris (3945)		
		TTL	OC	16	5	AM27S18AC	AMD	20	45 nsF	TTL	OC	16	5	AM27S20C	AMD		
						63S080	MMI (4026)							HM7610A-5	Harris (3944)		
						63S081	MMI (4026)							63S140	MMI (4026)		80
						AM27S19AC	AMD							29613A	† Raytheon (4131)		
						63S081	MMI (4026)							29613A	Raytheon		
	35 nsF	TTL	OC	16	5	AM27S18AM	† AMD	25	50 nsF	TTL	OC	16	5	HM7610B-2	† Harris (3944)		85
						53S080	† MMI (4026)							DM74S387	National (4057)		
						DM74S188	National (4057)							N82S126	Signetics		
						AM27S19AM	† AMD							HM7611B-2	† Harris (3944)		90
						53S081	† MMI (4026)							DM74S287	National (4057)		
						DM74S288	National (4057)							N82S129	Signetics		
	40 nsF	TTL	OC	16	5	AM27S18C	AMD	30	55 nsF	TTL	OC	16	5	HM7620B-1	† Harris (3945)		95
						TBP18SA30	TI							53S140	† MMI (4026)		
						AM27S19C	AMD							6300-1	MMI		
						TBP18S030	TI (542,4237)							63LS140	MMI		
	45 nsF	TTL	OC	16	5	DM54S188	† National (4057)	35						HM7621B-2	† Harris (3945)		100
						DM54S288	† National (4057)							53S141	† MMI (4026)		
	50 nsF	TTL	OC	16	5	AM27LS18C	AMD	40						6301-1	MMI		
						AM27S18M	† AMD							63LS141	MMI		
						HM7602-5	Harris (3943)							TBP24S10	TI (542,4238)		105
						6330-1	MMI							TBP24S10M	† TI (542,4238)		
						N82S23	Signetics							HM7611A-2	† Harris (3944)		110
						TBP18SA030M	† TI (542,4237)							TBP24S10M	† TI (542,4238)		
						AM27LS19C	AMD	45						HM7611A-2	† Harris (3944)		115
						AM27S19M	† AMD							IM5603C	Intersil		
						HM7603-5	Harris (3943)							S82S126	† Signetics (532)		
						6331-1	MMI							IM5623C	Intersil		
						N82S123	Signetics							S82S129	† Signetics (532)		
						TBP18S030M	† TI (542,4237)							HM7610-2	† Harris (3944)		120
	50 nsR	TTL	OC	16	5	SN54188A	† TI	50						5300-1	† MMI		
						SN74188A	TI							53LS140	† MMI		
	60 nsF	TTL	OC	16	5	HM7602-2	† Harris (3943)							TBP247A10M	† TI		
						5330-1	† MMI							HM7611-2	† Harris (3944)		125
						HM7603-2	† Harris (3943)							5301-1	† MMI		
						5331-1	† MMI							53LS141	† MMI		
	65 nsF	TTL	—	16	5	IM5610C	Intersil	55						IM5603M	† Intersil		
						IM5600C	Intersil							IM5623M	† Intersil		
						S82S23	† Signetics (532)										
						AM27LS19M	† AMD										
						S82S123	† Signetics (532)										
	75 nsF	TTL	—	16	5	IM5610M	† Intersil	60									
64x8	75 nsR	TTL	OC	24	5	SN74186	TI		80 nsF	TTL	OC	16	5				

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MEMORY-PROMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line					
<b>PROMs</b>									<b>(Cont'd)</b>													
256x8	1 μs *	Erasable	TS	24	5	CDP18042CD	RCA		512x4	70 nsF	TTL	TS	16	5	HM7621-5	Harris	(3945)	65				
	1 μsF	Erasable	TS	24	-9.5	AM1702A	AMD			HM7621A-2	† Harris	(3945)										
						AM1702AL	AMD			MCM7621	† Motorola	(4047)										
	1.5 μsF	Erasable	TS	24	-9.5	AM1702A-6	AMD			29611	† Raytheon											
	2.3 μsF	Erasable	TS	24	-9.5	MM8702A-4	National			29611	Raytheon											
	35 nsF	NMOS	OC	20	5	MB7117H	Fujitsu			S82S131	† Signetics	(532)								70		
				TS	20	5	MB7118H	Fujitsu			5305-1	† MMI										
	45 nsF	NMOS	OC	20	5	MB7117E	Fujitsu			53LS240	† MMI											
				TS	20	5	MB7118E	Fujitsu			5306-1	† MMI										
	60 nsF	TTL	TS	20	5	DM74LS471	National	(4057)		53LS241	† MMI										75	
	70 nsF	TTL	OC	20	5	6308-1	MMI			29613M	† Raytheon	(4131)										
						TBP28LA22	TI	(4240)		IM5624C	Intersil											
				24	5	6335-1	† MMI			IM5604C	Intersil											
				TS	20	5	6309-1	MMI			HM7620-2	† Harris	(3945)									
						DM54LS471	† National	(4057)		HM7621-2	† Harris	(3945)										
					TBP28L22	TI	(542,4240)	IM5624M	† Intersil													
			24	5	6336-1	MMI		IM5604M	† Intersil													
80 nsF	TTL	OC	20	5	5308-1	† MMI																
			TS	20	5	5309-1	† MMI															
					TBP28LA22M	† TI	(542,4240)															
550 nsF	Erasable	TS	24	-9.5	AM1702A-1	AMD																
					AM1702AL-1	AMD																
650 nsF	Erasable	TS	24	-9.5	AM1702A-2	AMD																
					AM1702AL-2	AMD																
512x4	30 nsF	TTL	OC	16	5	AM27S12AC	AMD		512x8	20 ns *	Reg	TS	24	5	TBP28R45	TI						
			TS	16	5	AM27S13AC	AMD			20 nsF	Reg	TS	24	5	AM27S25AC	AMD						
	40 nsF	TTL	OC	16	5	AM27S12AM	† AMD			25 nsF	Reg	TS	24	5	AM27S25AM	† AMD						
			TS	16	5	AM27S13AM	† AMD			27 nsF	Reg	TS	22	5	AM27S27C	AMD					85	
						20611A	Raytheon	(4131)		24	5	AM27S25C	AMD									
	45 nsF	TTL	OC	16	5	63S240	MMI	(4026)		30 nsF	Reg	TS	22	5	AM27S27M	† AMD						
			TS	16	5	63S241	MMI	(4026)		24	5	AM27S25M	† AMD									
	50 nsF	TTL	OC	16	5	AM27S12C	AMD			35 ns *	TTL	TS	20	5	TBP28P42	TI						
						HM7620A-5	Harris	(3945)		24	5	TBP28P45	TI									
						N82S130	Signetics					TBP28S45	TI									
				TS	16	5	AM27S13C	AMD			35 nsF	TTL	OC	20	5	AM27S28AC	AMD					
						HM7621A-5	Harris	(3945)					24	5	MB7123H	Fujitsu						
						MCM7621A	Motorola	(4047)														
				24	5	MCM7641A	Motorola	(4047)														
	55 nsF	TTL	OC	16	5	53S240	† MMI	(4026)														
					DM74S570	National	(4057)															
					DM74S570A	National	(4057)															
			TS	16	5	53S241	† MMI	(4026)														
					DM74S571	National	(4057)															
					DM74S571A	National	(4057)															
					DM74S571B	National	(4057)															
					29611	† Raytheon	(4131)															
					29611	Raytheon																
60 nsF	TTL	OC	16	5	AM27S12M	† AMD																
					6305-1	MMI																
					63LS240	MMI																
			TS	16	5	AM27S13M	† AMD															
					6306-1	MMI																
					63LS241	MMI																
					29613C	Raytheon	(4131)															
65 nsF	TTL	OC	16	5	DM54S570	† National	(4057)															
					DM54S570A	† National	(4057)															
			TS	16	5	DM54S571	† National	(4057)														
					DM54S571A	† National	(4057)															
					DM54S571B	† National	(4057)															
70 nsF	TTL	OC	16	5	HM7620-5	Harris	(3945)															
					HM7620A-2	† Harris	(3945)															
					S82S130	† Signetics	(532)															
							(Continued)															

† Military Temperature Range (-55° to 125°C) ns\*—Nanoseconds Typical nsF—Nanoseconds over Full Temperature Range nsR—Nanoseconds at Room Temperature  
 OC—Open Collector TS—Three-State OE—Open Emitter

MEMORY-PROMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
<b>PROMs</b>									<b>(Cont'd)</b>									
512x8	60 nsF	TTL	OC	24	5	HM7640A-2 † Harris (3946) DM54S475A † National (4057)			512x8	450 nsF	CMOS Erasable TS	24	5	IM665A-1	Intersil		60	
				20	5	HM7649-5 Harris (3947) DM54S472 † National (4057) DM54S472A † National (4057) DM54S472B † National (4057) DM54S474A † National (4057) DM54S474B † National (4057) DM74S472 † National (4057) TBP28S42 TI (542.4244)		5		550 nsF	CMOS Erasable TS	24	5	IM6654	Intersil			
				24	5	AM27S15C AMD HM7641A-2 † Harris (3946) N82S115 Signetics N82S141 Signetics TBP28S46 TI (542.543.4245)		10		600 nsF	CMOS Erasable TS	24	5	IM6654M † Intersil				
				20	5	AM27S28M † AMD DM54S473 † National (4057) TBP28S442 TI (4245)		15		35 nsF	TTL TS	20	5	DM74S472B † National (4057) DM77S8474 † National (4057)				
				24	5	DM54S475 † National (4057) DM74S475 † National (4057) TBP28S446 TI (4245)		20		40 nsF	TTL Reg	24	5	AM27S65 AMD (3902) DM54S572A † National (4057) DM54S573A † National (4057)			65	
65 nsF	TTL	OC	20	5	AM27S29M † AMD AM27S31M † AMD DM74S474 † National (4057)		25			35 nsF	TTL	OC	18	5	AM27S32AC AMD AM27S33AC AMD MB7121H Fujitsu		70	
				24	5	6348-1 MMI AM27S30M † AMD HM7640-5 Harris (3946) 6340-1 MMI		30						18	5	AM27S33AC AMD MB7122H Fujitsu 63S441A MMI (4026) DM74S573B † National (4057) N82S137B Signetics (4172)		75
				20	5	6349-1 MMI MCM7649 Motorola (4047) 29623 Raytheon (4131) TBP28S42M † TI TBP28S42M TI		35		35 nsR	TTL	OC	18	5	TBP24S441 TI (4239) TBP24S441M † TI (4239)			
				24	5	HM7641-5 Harris (3946) 5341-2 † MMI 6341-1 MMI MCM7641 Motorola (4047) DM54S474 † National (4057) TBP28S46M † TI (542.543.4245)		40						18	5	TBP24S41 TI (542.4239) TBP24S41M † TI (542.4239)		80
70 nsF	TTL	OC	20	5	5348-1 † MMI 5340-1 † MMI		45			40 nsF	TTL	OC	18	5	93452C Fairchild 93453C Fairchild			
				24	5	HM7649-2 Harris (3947) 5349-1 † MMI		50						18	5	AM27S32AM † AMD MB7121E Fujitsu HM7642B-5 Harris (3948) DM74S572A † National (4057)		85
				24	5	5341-1 † MMI		55						24	5	AM27S33AM † AMD MB7122E Fujitsu HM7643B-5 Harris (3948) 63S441 MMI (4026) DM74S573A † National (4057)		90
80 nsF	TTL	OC	20	5	TBP28S42M † TI (4245) HM7640-2 † Harris (3946) TBP28S446M † TI (4245)		60			45 nsF	TTL	OC	18	5	AM27S32C AMD 93452M † Fairchild HM7642B-2 † Harris (3948) HN25044 Hitachi			
				20	5	TBP28S42M † TI (542.4244) TBP28S42M TI		65						18	5	AM27S33C AMD 93453M † Fairchild HM7643B-2 † Harris (3948) HN25045 Hitachi 53S441 † MMI (4026) 63S3-2 MMI		100
				24	5	HM7641-2 † Harris (3946)		70						18	5	53RA441 † MMI (4030) HM7642A-5 Harris (3948) HM7643A-5 Harris (3948) 53S441A † MMI (4026) MCM7643A Motorola (4047) DM54S573B † National (4057)		105
85 nsF	TTL	OC	20	5	5340-2 † MMI AM27S15M † AMD S82S115 † Signetics (532) S82S141 † Signetics (532)		75			50 nsF	Reg TS TTL OC	18 18	5 5	53RA441 † MMI (4030) HM7642-2 † Harris (3948) HM7642-5 Harris (3948) 6350-1 MMI 6352-1 MMI DM74S572 † National (4057)				
				20	5	HM6641-2 † Harris (3969) HM6641-8 † Harris (3969) HM6641-9 Harris (3969)		80						18	5	HM7643-2 † Harris HM7643-5 Harris (3948) 6351-1 MMI 6353-1 MMI DM74S573 † National (4057) N82S137 Signetics (4172)		110
				24	5	IM6654A Intersil X2804 Xicor (4267.4268)		85						18	5	AM27S32M † AMD AM27S33M † AMD DM77S181A † National (4057)		115
90 nsF	TTL	OC	24	5				90		65 nsF	TTL	OC	18	5				120
				24	5			95						24	5			
				24	5			100										
				24	5			105										
				24	5			110										
				24	5			115										
				24	5			120										

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



# IC MASTER

## MEMORY-PROMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line		
<b>PROMs</b>									<b>(Cont'd)</b>										
1024x8	450 nsF	Erasable	TS	24	±5,12				2048x4										
										95 nsF	TTL	TS	18	5	29653M	† Raytheon	(4131)		
						MCM68708	Motorola			100 nsF	TTL	OC	18	5	5388-1	† MMI			
						INS8708	National					TS	18	5	5389-1	† MMI	65		
						<b>TMS2708-45</b>	TI	<b>(543,4234)</b>							N82S185	Signetics			
						<b>TMS27L08-45</b>	TI	<b>(543,4234)</b>			115 nsF	TTL	TS	18	5	<b>S82S185</b>	† Signetics	<b>(532)</b>	
2048x4	—	Reg	TS	24	5	AM27S75	AMD	<b>(3902)</b>	5	450 nsF	CMOS Erasable				IM6657	Intersil			
		TTL	TS	20	5	TBP28R165	TI					TS	24	5	IM6657M	† Intersil			
	35 nsF					AM27S184AC	AMD			2048x8	—	TTL	TS	24	5	AM27S45	AMD	<b>(3904)</b>	
						MB7127Y	Fujitsu									AM27S45A	AMD	<b>(3904)</b>	
						AM27S185AC	AMD		10							AM27S47	AMD	<b>(3904)</b>	
						MB7128Y	Fujitsu									AM27S47A	AMD	<b>(3904)</b>	
						63S841A	MMI	<b>(4026)</b>		20 ns *	Reg	TS	24	5	TBP28R166	† TI			
						<b>DM87S185B</b>	National	<b>(4057)</b>		35 ns *	TTL	OC	24	5	TBP28SA166	TI			
	45 ns *	TTL	OC	18	5	SN54S455	† TI		15			TS	24	5	TBP28P166	† TI			
						SN74S455	TI								<b>TBP28S166</b>	TI	<b>(4247)</b>		
						<b>TBP24SA81</b>	TI	<b>(4240)</b>		35 nsF	Reg	TS	24	5	63RA1681A	MMI	<b>(4035)</b>		
						<b>TBP24SA81M</b>	† TI	<b>(4240)</b>							63RS1681A	MMI	<b>(4038)</b>		
						<b>TBP24S81</b>	TI	<b>(542,4239)</b>							MCM27S45	Motorola			
						<b>TBP24S81M</b>	† TI	<b>(542,4239)</b>							MCM27S47	Motorola			
45 nsF	TTL	OC	18	5		AM27S184AM	† AMD		20		TTL	OC	24	5	AM27S190AC	AMD			
						MB7127H	Fujitsu								AM27S290AC	AMD			
						AM27S185AM	† AMD								AM27S191A	AMD			
						MB7128H	Fujitsu								AM27S291A	AMD			
						<b>DM87S185A</b>	National	<b>(4057)</b>							MB7138YZ	Fujitsu			
						<b>N82S185B</b>	Signetics	<b>(4178)</b>							3636B-1	Intel			
50 nsF	TTL	OC	18	5		AM27S184C	AMD		25						<b>63S1681A</b>	MMI	<b>(4026)</b>		
						AM27S185C	AMD								MCM27S191	Motorola			
						<b>HM7685A-5</b>	Harris	<b>(3950)</b>							MCM27S291	Motorola			
						53S841A	† MMI	<b>(4026)</b>							<b>DM87S191B</b>	National	<b>(4057)</b>		
						63S841	MMI	<b>(4026)</b>							<b>DM87S291B</b>	National	<b>(4057)</b>		
						<b>DM77S185B</b>	† National	<b>(4057)</b>		40 nsF	Reg	TS	24	5	53RA1681A	† MMI	<b>(4035)</b>		
						<b>N82S185A</b>	Signetics	<b>(4178)</b>							53BS1681A	† MMI	<b>(4038)</b>		
55 nsF	TTL	OC	18	5		AM27S184M	† AMD		30						63RA1681	MMI	<b>(4035)</b>		
						<b>DM87S184</b>	National	<b>(4057)</b>							63RS1681	MMI	<b>(4038)</b>		
						AM27S185M	† AMD			45 nsF	Reg	TS	24	5	53RA1681	† MMI	<b>(4035)</b>		
						MB7128E	Fujitsu		35						53RS1681	† MMI	<b>(4038)</b>		
						MB7128E-W	† Fujitsu								932510C	Fairchild			
						53S841	† MMI	<b>(4026)</b>							932510C	Fairchild			
						6389-2	MMI								MB7137H	Fujitsu			
						<b>DM87S185</b>	National	<b>(4057)</b>							932511C	Fairchild			
60 nsF	TTL	TS	18	5		AM27LS185C	AMD		40						932511C	Fairchild			
						AM27PS185C	AMD								MB7138H	Fujitsu			
						<b>HM7685A-2</b>	† Harris	<b>(3950)</b>							3636B-2	Intel			
						<b>DM77S185A</b>	† National	<b>(4057)</b>							<b>DM87S191A</b>	National	<b>(4057)</b>		
						29651A	Raytheon	<b>(4131)</b>							<b>DM87S291A</b>	National	<b>(4057)</b>		
						29653A	Raytheon	<b>(4131)</b>		50 nsF	TTL	OC	24	5	AM27S190C	AMD			
									45						AM27S290AM	† AMD			
															AM27S290C	AMD			
															HN24084S	Hitachi			
65 nsF	TTL	OC	18	5		AM27LS184M	† AMD		50						HN25084S	Hitachi			
						AM27LS185M	† AMD								<b>TBP28SA86A-50</b>	TI	<b>(4247)</b>		
						AM27PS185M	† AMD								AM27PS191C	AMD			
70 nsF	TTL	OC	18	5		6388-1	MMI								AM27PS291C	AMD			
						<b>DM77S184</b>	† National	<b>(4057)</b>							AM27S191	AMD			
															AM27S191AM	† AMD			
						<b>HM7685-5</b>	Harris	<b>(3950)</b>							AM27S291	AMD			
						5389-2	† MMI								AM27S291AM	† AMD			
						6389-1	MMI								<b>HM76161A-5</b>	Harris	<b>(3951)</b>		
						<b>MCM7685</b>	Motorola	<b>(4047)</b>	55						HN25085S	Hitachi			
						<b>DM77S185</b>	† National	<b>(4057)</b>							3636B	Intel			
						29651C	Raytheon	<b>(4131)</b>							82HS191	Intel			
															<b>53S1681A</b>	† MMI	<b>(4026)</b>		
75 nsF	TTL	TS	18	5		29653C	Raytheon	<b>(4131)</b>							63S1681	MMI	<b>(4026)</b>		
80 nsF	TTL	TS	18	5		<b>S82S185A</b>	Signetics	<b>(532,4178)</b>							<b>MCM76161A</b>	Motorola	<b>(4047)</b>		
									60						<b>DM77S191B</b>	† National	<b>(4057)</b>		
															<b>DM77S291B</b>	† National	<b>(4057)</b>		
						<b>S82S191A</b>	† Signetics	<b>(532,4180)</b>											
90 nsF	TTL	TS	18	5		<b>HM7685-2</b>	† Harris	<b>(3950)</b>											
						29651M	† Raytheon	<b>(4131)</b>											

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.





MEMORY-PROMs (Cont'd)

Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line			
<b>PROMs</b>								<b>(Cont'd)</b>										
4096x8	50 nsF	TTL	TS	24	5	M3632 M3632	† Intel Intel	4096x8	550 nsF	Erasable	TS	24	5	NMC27C32-55	National	(Cont'd)		
	55 nsF	TTL	OC TS	24 24	5 5	MB7141H MB7142H DM87S321 DM87S421	Fujitsu Fujitsu National (4057) National (4057)	5	8192x8	45 nsF	TTL	OC TS	24 24	5 5	93Z564 93Z565	Fairchild Fairchild	70	
	65 nsF	TTL	OC TS	24 24	5 5	MB7141E MB7142E HM76321-5 DM77S321 DM77S421	Fujitsu Fujitsu Harris (3953) † National (4057) † National (4057)	10		55 nsF	TTL	OC TS	24 24	5 5	MB7143H MB7144H	Fujitsu Fujitsu	75	
	70 nsF	TTL	TS	24	5	29671A 29671A	† Raytheon (4131) Raytheon	15		85 nsF	TTL	TS	24	5	HM76641-5	Harris (3954)		
	80 nsF	TTL	TS	24	5	29671 29671 29671A 29671A N82S321	† Raytheon (4131) Raytheon † Raytheon Raytheon Signetics	20		100 nsF	TTL	TS	24	5	HM76641-2	† Harris (3954)		
	85 nsF	TTL	TS	24	5	HM76321-2 29673 29673	† Harris (3953) † Raytheon (4131) Raytheon	25		175 nsF	CMOS	TS	28	5	HM6664-2 HM6664-8 HM6664-9	† Harris (3969) † Harris (3969) Harris (3969)		
	100 nsF	TTL	TS	24	5	29671 29671	† Raytheon Raytheon	30		200 nsF	CMOS, Erasable	TS	28	5	HN27C64-20 AM2764-20 MBM2764-20 HN26C64-20 2764-2 5133-2 5133-200 TMM2764-2	Hitachi (3974) AMD (3928) Fujitsu Hitachi Intel SEEQ (513.4137) SEEQ (513.4137) Toshiba	80	
	105 nsF	TTL	TS	24	5	29673 29673	† Raytheon Raytheon	35		250 nsF	CMOS, Erasable	TS	28	5	MBM27C64-25 HN27C64-25	Fujitsu Hitachi (3974)	85	
	150 nsF	Erasable	TS	24	5	AM2732A-1 AM2732A-15	AMD (3926) AMD (3926)	40		Erasable	TS	28	5	AM2764-25 2764 MBM2764-25 HN482764-3 2764 2764-25 5133-250 M5133-250	AMD (3928) Fairchild Fujitsu Hitachi Intel SEEQ (513.4137) † SEEQ (513)	90		
	200 nsF	Erasable	TS	24	5	AM2732A-2 AM2732A-20 MBM2732A-20 HN482732A-20 2732A-2 2732A-20	AMD (3926) AMD (3926) Fujitsu Hitachi Intel Intel	45		300 nsF	CMOS, Erasable	TS	28	5	MBM27C64-30 MBM27C64-30W HN27C64-30	Fujitsu Fujitsu Hitachi (3974)	95	
	250 nsF	Erasable	TS	24	5	AM2732A AM2732A-25 AM2732A-25 MBM2732A-25 HN482732A-25 2732A 2732A-25 TMM2732-2	AMD (3926) AMD (3926) AMD Fujitsu Hitachi Intel Intel Toshiba	50		Erasable	TS	28	5	AM2764-30 MBM2764-30 MBM2764-30X 2764-3 2764-30 5133-3 5133-300	AMD (3928) Fujitsu Fujitsu Intel Intel SEEQ (513.4137) SEEQ (513.4137)	100		
	300 nsF	Erasable	TS	24	5	AM2732A-3 AM2732A-30 MBM2732A-30 MBM2732A-30X HN472632A-30 HN482732A-30 2732A-3 2732A-30	AMD (3926) AMD (3926) Fujitsu Fujitsu Hitachi Hitachi Intel Intel	55		350 nsF	Erasable	TS	24	5	MCN68766-35 TMS2564-35	Motrola (4047) TI (543.4229)	105	
	350 nsF	Erasable	TS	24	5	RO9433C	GI	60		400 nsF	Erasable	TS	28	5	5133-4	SEEQ (513.4137)		
	450 nsF	Erasable	TS	24	5	NMC27C32-35 M58735	National Mitsubishi	65		450 nsF	Erasable	TS	24	5	MCN68764 MCN68766	Motrola (4047) Motrola (4047)		
		NMOS	TS	24	5	AM2732A-4 AM2732A-45	AMD (3926) AMD (3926)			500 nsF	Erasable	TS	28	5	AM2764-45 HN482764-4 2764-4 2764-45 M5L2764 MSM2764A 5133-450 M5133-450 TMS2564-45	AMD (3928) Hitachi Intel Intel Mitsubishi OKI (4118) SEEQ (513.4137) SEEQ (513) TI (543.4229)	110	
		NMOS	TS	24	5	AM2732A-45 F2732 HN462532 HN462732 2732A-4 M5L2732 NMC27C32-45 μPD2732 TMM2732	AMD (3926) Fairchild Hitachi Hitachi Intel Mitsubishi National NEC Toshiba			16384x8	150 nsF	Erasable	TS	28	5	AM27128-1 AM27128-15	AMD (3929) AMD (3929)	120
		NMOS	TS	24	5	RO94132	GI				200 nsF	Erasable	TS	28	5	AM27128-2 AM27128-20 AM27128-20 M27128 5143-2 5143-200	AMD (3929) AMD (3929) AMD † Intel SEEQ (513) SEEQ (513)	125
											250 nsF	Erasable	OE TS	28 28	5 5	HM4827128-25 AM27128	Hitachi (3975) AMD (3929)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MEMORY-PROMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line					
<b>PROMs</b>						<b>(Cont'd)</b>							
16384x8	250 nsF	Erasable	TS	28	5	(Cont'd)							
						AM27128-25 †AMD		(3929)					
						AM27128-25	AMD						
						MBM27128-25	Fujitsu						
						HN4827128-25	Hitachi						
						27128-25	Intel	5					
						5143-250	SEEQ	(513)					
						M5143-250 †SEEQ		(513)					
						300 nsF	Erasable	OE	28	5	HM4827128-30	Hitachi	(3975)
											27128-30	Intel	
TS	28	5	AM27128-3	AMD	(3929)								
			AM27128-30	AMD	(3929)								
			MBM27128-30	Fujitsu									
			HN4827128-30	Hitachi									
			27128-3	Intel									
			5143-3	SEEQ	(513)								
			5143-300	SEEQ	(513)								
			NMOS	TS	28						5	MBM27128-30W	
						† Fujitsu							
			450 nsF	Erasable	OE	28	5	HM4827128-45	Hitachi	(3975)			
27128-45	Intel												
M27128-45 †	Intel	20											
TS	28	5						AM27128-4	AMD	(3929)			
								AM27128-45 †AMD		(3929)			
								AM27128-45	AMD				
								HN4827128-45	Hitachi				
								27128-4	Intel				
								5143-4	SEEQ	(513)			
								5143-450	SEEQ	(513)			
			M5143-450 †SEEQ		(513)								
			32768x8	170 nsF	Erasable	TS	28	5	AM27256-1	AMD	(3930)		
									MBM27C256-20	Fujitsu			
200 nsF	CMOS	TS		28	5	AM27256-2	AMD	(3930)					
						AM27256-20	AMD	(3930)					
NMOS	TS	28		5	R09256D	GI							
					250 nsF	CMOS	TS	28	5	MBM27C256-25	Fujitsu		
AM27256	AMD	(3930)											
AM27256-25	AMD	(3930)											
27256	Intel	(4025)											
NMOS	TS	28		5	R09256DS	GI							
			300 nsF		CMOS	TS	28	5	MBM27C256-30	Fujitsu			
AM27256-3	AMD	(3930)											
AM27256-30	AMD	(3930)											
R09256CS	GI												
350 nsF	NMOS	TS	28	5	AM27256-4	AMD	(3930)						
								450 nsF	Erasable	TS	28	5	AM27256-45
NMOS	TS	28	5	R09256B	GI	45							

† Military Temperature Range (-55° to 125°C)  
OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range  
TS—Three-State

nsR—Nanoseconds at Room Temperature  
OE—Open Emitter

MEMORY-RAMs

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
<b>Bubble</b>									8192x8	120 nsF	NMOS	TS	28	5	IMS2630-12	Immes	(3980)	50
Note: Bubble Memory Support Circuits (controllers, function drivers, coil drivers, sense amplifiers), see Interface, Memory and Peripheral Drivers; Interface—Sense Amplifiers; Linear-Other Devices (Volume II).									150 nsF	NMOS	TS	28	5	IMS2630-15	Immes	(3980)		
1 Megabit (2048x512)									200 nsF	NMOS	TS	28	5	IMS2630-20	Immes	(3980)		
1 Megabit (4096x256x1)									250 nsF	NMOS	TS	28	5	2186-25	Intel			
4 Megabit (512x8192)									300 nsF	NMOS	TS	28	5	2186-30	Intel		55	
74032-Bits Serial									350 nsF	NMOS	TS	28	5	2187A-30	Intel			
74032-Bits Serial									370 ms	NMOS	—	18		2187A-35	Intel		60	
83354-Bits Serial									4.5 ms	NMOS	—	18		2186-35	Intel			
262144 (1024x256)														2187A-25	Intel		65	
273745-Bits Serial									6.0 ms			20		MSM2167-55	Mitsubishi			
283026-Bits Serial									8.5 ms	NMOS	—	16		MSM2167-70	Mitsubishi		70	
296128-Bits Serial									370 ms	Card				MB8117-10	Fujitsu			
324024-Bits Serial									4.5 ms	Card				MB8118-10	Fujitsu			
<b>CCD</b>														HM8416A-3	Hitachi		75	
32768-Bits (4096x8x1)									TS	16	± 5,12	F232	Fairchild					
65535-Bits (4096x16x1)									TS	16	± 5,12	F264	Fairchild					
<b>Dynamic</b>														MK4516-10	Mostek		80	
2048x8									50 nsF	NMOS	OC	24	5	μPB409	NEC			
4096x1									120 nsF	NMOS	TS	16	± 5,12	μPB429	NEC		85	
135 nsF									NMOS	TS	22	± 5,15	μPD411-4	NEC				
150 nsF									NMOS	TS	16	± 5,12	MK4027-2	Intersil		90		
200 nsF									NMOS	TS	16	± 5,12	ITT4027-2	ITT				
250 nsF									NMOS	TS	16	± 5,12	MK4027-2	Mostek		95		
300 nsF									NMOS	TS	22	± 5,12	MCM4027A-2	Motorola				
350 nsF									NMOS	TS	16	± 5,12	M4027-2	SGS		100		
400 nsF									NMOS	TS	16	± 5,12	μPD411-3	NEC				
450 nsF									NMOS	TS	16	± 5,12	IM7027-3	Intersil		105		
500 nsF									NMOS	TS	16	± 5,12	MK4027-3	Intersil				
550 nsF									NMOS	TS	16	± 5,12	ITT4027-3	ITT		110		
600 nsF									NMOS	TS	16	± 5,12	MK4027-3	Mostek				
650 nsF									NMOS	TS	16	± 5,12	MKB4027-83	Mostek		115		
700 nsF									NMOS	TS	16	± 5,12	MCM4027A-3	Motorola				
750 nsF									NMOS	TS	16	± 5,12	M4027-3	SGS		120		
800 nsF									NMOS	TS	16	± 5,12	μPD411-2	NEC				
850 nsF									NMOS	TS	16	± 5,12	μPD411A-2	NEC		125		
900 nsF									NMOS	TS	16	± 5,12	IM7027-4	Intersil				
950 nsF									NMOS	TS	16	± 5,12	MK4027-4	Intersil		130		
1000 nsF									NMOS	TS	16	± 5,12	ITT4027-4	ITT				
1050 nsF									NMOS	TS	16	± 5,12	MK4027-4	Mostek		135		
1100 nsF									NMOS	TS	16	± 5,12	MKB4027-84	Mostek				
1150 nsF									NMOS	TS	16	± 5,12	MCM4027A-4	Motorola		140		
1200 nsF									NMOS	TS	16	± 5,12	M4027-4	SGS				
1250 nsF									NMOS	TS	22	± 5,12	μPD411-1	NEC		145		
1300 nsF									NMOS	TS	22	± 5,12	μPD411A-1	NEC				
1350 nsF									NMOS	TS	22	± 5,12	μPD411	NEC		150		
1400 nsF									NMOS	TS	22	± 5,12	μPD411A	NEC				
1450 nsF									NMOS	TS	16	± 5,12	ITT4027-6	ITT		155		
1500 nsF									NMOS	TS	16	± 5,12						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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Master Selection Guide

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MEMORY-RAMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
<b>Dynamic (Cont'd)</b>									<b>65536x1 150 nsF NMOS TS 16 5</b>							
16384x1	200 nsF	NMOS	TS	16	±5,12	TMS4116-20	TI	(Cont'd)						HM4864-2	Hitachi	(Cont'd)
						TMM416-3	Toshiba							HM4864A-15	Hitachi	
	235 nsF	NMOS	TS	16	5	MKB4516-80	† Mostek							IMS2600-15	Immes (3979)	70
	250 nsF	NMOS	TS	16	±5,12	AM9016D	AMD (3918)	5						2164A-15	Intel	
						F4116-4	Fairchild							MT4264-15A	MicronTech	
						HM4716A-4	Hitachi							M5K4164-15	Mitsubishi	
						ITT4116-4	ITT							MK4564-15	Mostek	
						MK4116-4	Mostek							MKB4564-82	† Mostek	
						MKB4116-84	† Mostek							MCN6664A-15	Motorola (4046)	75
						MCN4116B-25	Motorola (4046)	10						MCN6665A-15	Motorola (4046)	
						NM5290-4	National							MCN6665B-15	Motorola (4046)	
						μPD416-1	NEC (4073)							NMC4164-15	National	
						HYB4116-4	Siemens							μPD4164-15	NEC (4077)	
						TMS4116-25	TI							μPD4164-3	NEC (4077)	80
						TMM416-4	Toshiba	15						MSM3764-15	OKI (4118)	
	270 nsF	NMOS	TS	16	5	MKB4516-81	† Mostek							MM4164-15	Panasonic (4120)	
	300 nsF	NMOS	TS	16	±5,12	AM9016C	AMD (3918)							HYB4164-2	Siemens	
						MCN4116B-30	Motorola (4046)							SMLJ4164-15	† TI (4219)	85
						μPD416	NEC (4073)							TMS4164-15	TI (543,4219)	
														TMM4164-3	Toshiba	
	320 nsF	NMOS	TS	16	5	MKB4516-82	† Mostek	20						TMS4161	TI (4214)	
16384x4	100 nsF	NMOS	TS	18	5	MB81416-10	Fujitsu		200 nsF	NMOS	TS	16	5	F4164-20	Fairchild	
	120 nsF	NMOS	TS	18	5	MB81416-12	Fujitsu							F64K-20	Fairchild	90
	150 nsF	NMOS	TS	18	5	MB81416-15	Fujitsu							MB8264-20	Fujitsu	
						SMLJ4416-15	† TI (4220)							MB8265-20	Fujitsu	
						TMS4416-15	TI (4220)	25						HM4864-3	Hitachi	
	200 nsF	NMOS	TS	18	5	SMLJ4416-20	† TI (4220)							HM4864A-20	Hitachi	
						TMS4416-20	TI (4220)							HM4865A-20	Hitachi	
	250 nsF	NMOS	TS	18	5	TMS4416-25	TI (4220)	30						2164A-20	Intel	95
32768x1	200 nsF	NMOS	TS	18	±5,12	MK4332D-3	Mostek							M2164A-20	† Intel	
						MKM4332-83	† Mostek							MT4264-20A	MicronTech	
	250 nsF	NMOS	TS	18	±5,12	MKM4332-84	† Mostek							M5K4164-20	Mitsubishi	
64536x1	150 nsF	NMOS	TS	16	5	MB8264A-15W	† Fujitsu							MK4564-20	Mostek	
	200 nsF	NMOS	TS	16	5	MB8264A-20W	† Fujitsu							MKB4564-83	† Mostek	100
65536x1	55 nsF	NMOS	TS	16	5	MB8281	Fujitsu							MCN6664A-20	Motorola (4046)	
	100 nsF	NMOS	TS	16	5	F4164	Fairchild	35						MCN6665A-20	Motorola (4046)	
						MB8264A-10	Fujitsu							MCN6665B-20	Motorola (4046)	
						MB8265A-10	Fujitsu							NMC4164-20	National	
						MB8266A-10	Fujitsu							μPD4164-0	NEC (4077)	105
						IMS2600-10	Immes (3979)							μPD4164-20	NEC (4077)	
						μPD4164-2	NEC (4077)	40						MSM3764-20	OKI (4118)	
	120 nsF	NMOS	TS	16	5	F4164-12	Fairchild							MM4164-20	Panasonic (4120)	
						F64K-12	Fairchild							HYB4164-3	Siemens	
						MB8264A-12	Fujitsu							SMLJ4164-20	† TI (4219)	110
						MB8265A-12	Fujitsu							TMS4164-20	TI (543,4219)	
						MB8266A-12	Fujitsu							TMM4164-4	Toshiba	
						HM4864-1	Hitachi									
						HM4864A-12	Hitachi									
						HM4865A-12	Hitachi									
						IMS2600-12	Immes (3979)									
						MT4265-12A	MicronTech	50								
						MCN6665A-12	Motorola (4046)									
						MCN6665B-12	Motorola (4046)									
						NMC4164-12	National									
						μPD4164-12	NEC (4077)									
						MSM3764-12	OKI (4118)									
						SMLJ4164-12	† TI (4219)									
						TMS4164-12	TI (543,4219)									
						TMM4164-2	Toshiba									
	150 nsF	NMOS	TS	16	5	F4164-15	Fairchild									
						F64K-15	Fairchild									
						MB8264-15	Fujitsu									
						MB8264A-15	Fujitsu									
						MB8265-15	Fujitsu									
						MB8265A-15	Fujitsu									
						MB8266A-15	Fujitsu	60								
								65								

† Military Temperature Range (-55° to 125°C) ns\*—Nanoseconds Typical nsF—Nanoseconds over Full Temperature Range nsR—Nanoseconds at Room Temperature  
OC—Open Collector TS—Three-State OE—Open Emitter

MEMORY-RAMs (Cont'd)

Table with columns: Access Time (Max), Type, Output, No. Pins, Supply Voltage, Device, Source, Line. Sections include Dynamic, Static, 262144x1, 4x4, 4x8, 8x2, 16x1, 16x12, 16x4, 35 nsR, 40 ns\*, 45 nsF, 50 nsF, 60. Includes various device models and manufacturers like Fujitsu, Hitachi, Motorola, AMD, National, TI, Signetics.

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MEMORY-RAMs (Cont'd)

Organiza-tion	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
<b>Static</b>									
<b>(Cont'd)</b>									
	<b>16x4</b>	50 nsF	TTL	TS	16	5			
							54S189 † Fairchild DM54S189 † National DM8599 National 54S189 † Signetics (532) SN54S189A † TI (988)		
				28	5	IDM2970C	National	5	
	53 nsF	TTL	TS	28	5	AM29705C	AMD (1425)		
						F29705M † National IDM29705 † National			
	55 ns *	TTL	OC	16	5	54LS289 † Fairchild 54LS89 † Fairchild		10	
				TS	16	54LS189 † Fairchild			
	55 nsF	TTL	OC	16	5	AM27LS02C	AMD		
						AM27LS06C	AMD		
				28	5	IDM29704M † National		15	
				TS	16	5	AM27LS03C	AMD	
						AM27LS07C	AMD		
	58 nsF	TTL	TS	28	5	AM29705M † AMD	(1425)		
	60 nsF	TTL	OC	16	5	AM27LS06M † AMD AM3101 AMD AM5489 † AMD DM8589 National μPB2089 NEC S82S25 † Signetics SN54LS289A † TI (1012) SN54LS319A † TI (1017) SN74LS289A TI (1012) SN74LS319A TI (1017)		20	
				TS	16	5	AM27LS07M † AMD SN54LS189A † TI (988) SN54LS219A † TI (994) SN74LS189A TI (988) SN74LS219A TI (994)		25
	60 nsR	TTL	OC	16	5	HD7489 Hitachi SN7489 TI		30	
	70 nsF	TTL	OC	16	5	AM27LS02C	AMD		
				TS	16	5	AM27LS03C	AMD	
						DM7599 † National		35	
	80 nsF	TTL	OC	16	5	DM7589 † National			
	85 nsF	TTL	OC	16	5	AM27LS02M † AMD AM27LS03M † AMD		40	
	110 nsR	TTL	OC	16	5	AM31L01C	AMD		
						AM31L01M † AMD			
	120 nsR	CMOS	TS	18	4.5-12.5	F4710BC	Fairchild		
						F4710BM † Fairchild		45	
	650 nsR	CMOS	TS	16	5	MM74C89 National CD40114B † RCA CD4011BE		50	
							(837)		
	16x9	40 ns *	TTL	OC	20	5	SN54LS311 † TI SN54LS312 † TI SN74LS311 TI SN74LS312 TI		
				TS	20	5	SN54LS211 † TI SN54LS212 † TI SN74LS211 TI SN74LS212 TI		
	16x12	40 ns *	TTL	TS	20	5	SN54LS213 † TI SN74LS213 TI		
	32x2	50 nsF	TTL	OC	16	5	N82S21	Signetics	
	32x8	40 ns *	TTL	OC	20	5	SN54LS318 † TI SN74LS318 TI		60
				TS	20	5	SN54LS218 † TI SN74LS218 TI		

(Continued)

Organiza-tion	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
<b>32x8</b>									
	300 ns *	CMOS	TS	18	10	HCMP1824 Hughes HMMP1824 † Hughes CDP1824 † RCA		65	
							(1594)		
	600 ns *	CMOS	TS	18	5	CDP1824C † RCA HCMP1824C Hughes HMMP1824C † Hughes		70	
							(1594)		
	64x1	1.05 μsR	CMOS	TS	24	5-15	MCM14552A † Motorola HD14552B Hitachi MCM14552C Motorola		
		2.1 μsR	CMOS	TS	24	5-15			
	10 nsR	ECL	OE	16	-5.2	SN10142	TI		
	12 nsF	ECL	OE	16	-5.2	μPB10142	NEC		
	15 nsF	ECL	OE	16	-5.2	HD10148 Hitachi MCM10148 † Motorola (4046) μPB10148		75	
							NEC		
	15 nsR	ECL	OE	16	-5.2	SN10140 TI SN10148 TI			
	270 nsR	CMOS	—	14	3-18	MCM14505A † Motorola HEF4505 † Signetics		80	
							(864)		
	350 nsR	CMOS	—	14	4.5-16	HD14505B Hitachi MCM14505C Motorola			
	64x4	6 ns *	ECL	OE	24	-5.2	GXB100473 Siemens		
		40 ns *	TTL	OC	16	5	SN54LS316 † TI SN54LS317 † TI SN74LS316 TI SN74LS317 TI		85
				TS	16	5	SN54LS216 † TI SN54LS217 † TI SN74LS216 TI SN74LS217 TI		90
	2000 ns *	CMOS	TS	16	5	MN1203	Panasonic		
	64x8	800 nsF	CMOS	TS	22	5	CDP1826C	RCA (1594)	
	64x9	35 nsF	TTL	OC	28	5	93419A Fairchild N82S09 Signetics N82S19 Signetics		95
				TS	28	5	93419C Fairchild MBM93419 Fujitsu		100
	60 nsF	TTL	OC	28	5	93419M † Fairchild S82S19 † Signetics (532)			
	80 nsF	TTL	OC	28	5	S82S09 † Signetics (532)			
	64x12	150 nsF	CMOS	TS	18	3-11	IM6512AI Intersil IM6512AM † Intersil		105
		460 nsF	CMOS	TS	18	3-7	IM6512I Intersil IM6512M † Intersil		
		600 nsF	CMOS	TS	18	3-7	IM6512C Intersil		
	128x1	12 nsR	ECL	OE	16	-5.2	MCM10147 † Motorola (4046) GXB10147A Siemens		110
		15 nsR	ECL	OE	16	-5.2	HD10147 Hitachi SN10147 TI		
	128x8	—	CMOS	—	8	5	PCD8571 Signetics S68B10 AMI F68B10 Fairchild MCM68B10 Motorola (4046)		115
		250 ns *	NMOS	TS	24	5			
				TS	24	10	CDP1823 RCA (1594) EF68B10 Thomson-CSF		
		250 nsF	CMOS	TS	24	5	S6810-1 AMI		
		350 nsF	NMOS	TS	24	5	S68A10 AMI F68A10 Fairchild HM468A10 Hitachi MCM68A10 Motorola (4046) EF68A10 Thomson-CSF		120
		360 nsF	NMOS	TS	24	5			
		450 nsF	CMOS	TS	24	5	CDP1823C RCA (1594) S6810 AMI F6810 Fairchild (1461) HM46810 Hitachi		125
			NMOS	TS	24	5			

(Continued)

† Military Temperature Range (-55° to 125°C)  
OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range  
TS—Three-State

nsR—Nanoseconds at Room Temperature  
OE—Open Emitter

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MEMORY-RAMs (Cont'd)

Table with columns: Access Time (Max), Type, Output, No. Pins, Supply Voltage, Device, Source, Line. It is divided into two main sections: Static and 256x1, 256x4. The Static section lists various RAM types like MCM6810, MCM14537A, HD14537B, etc. The 256x1 and 256x4 sections list RAMs like CO40061, MK4007, ITT5101S, etc. The table includes device names, manufacturers, and typical values.

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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MEMORY-RAMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line							
Static								(Cont'd)							
256x4								(Cont'd)							
	200 nsF	NMOS	TS	16	5	SY21H12-2	Synertek								
				18	5	SY21H11-2	Synertek								
				22	5	SY21H01-2	Synertek								
	220 nsF	CMOS	TS	18	5	HM6561B-2	Harris (3965)	5							
						HM6561B-8	Harris (3965)								
						HM6561B-9	Harris (3965)								
						IM65X61-1	Intel								
						IM65X61-1	Intersil								
						IM65X61-1M	Intersil								
				22	5	HM6551B-2	Harris (3965)	10							
						HM6551B-8	Harris (3965)								
						HM6551B-9	Harris (3965)								
						IM65X51-1	Intersil								
						IM65X51-1M	Intersil								
	235 nsF	CMOS	TS	18	10	IM65X61A	Intersil	15							
						IM65X61AM	Intersil								
				22	10	IM65X51A	Intersil								
						IM65X51AM	Intersil								
	240 nsF	CMOS	TS	22	5	H66551RH	Harris (4689)	20							
	250 nsF	CMOS	TS	22	5	NWS5101EL2	RCA (1594)								
						SCM5101-1B	SSS								
						CDP1822	RCA (1594)								
						TCC-244	RCA								
		NMOS	TS	16	5	AM9101D	AMD	25							
						AM9111D	AMD								
						AM9112D	AMD								
						SY2112A-2	Synertek (4210)								
				18	5	μPD2111AL-2	NEC								
						SY2111A-2	Synertek (4210)								
				22	5	SY2101A-2	Synertek (4210)	30							
	300 nsF	CMOS	TS	18	5	HM6561-2	Harris (3965)								
						HM6561-8	Harris (3965)								
						HM6561-9	Harris (3965)								
						IM65X61	Intersil								
						IM65X61M	Intersil								
				22	5	HM6551-2	Harris (3965)	35							
						HM6551-8	Harris (3965)								
						HM6551-9	Harris (3965)								
						IM65X51	Intersil								
						IM65X51M	Intersil								
						LH-5101S	Sharp	40							
		NMOS	TS	16	5	AM9101C	AMD	45							
						AM9101CM	AMD								
						AM9111C	AMD								
						AM9111CM	AMD								
						AM9112C	AMD								
						AM9112CM	AMD								
						AM91L01C	AMD								
						AM91L01CM	AMD								
						AM91L11C	AMD								
						AM91L11CM	AMD								
						AM91L12C	AMD								
						AM91L12CM	AMD								
													AM91L12C	AMD	
	350 nsF	CMOS	TS	18	5	HM6561-5	Harris (3965)	55							
						IM65X61C	Intersil								
				22	5	HM6551-5	Harris (3965)								
						IM65X51C	Intersil								
						NWS5101DL3	RCA (1594)								
						NWS5101EL3	RCA (1594)								
						SCM5101-1A	SSS								
		NMOS	TS	16	5	SY2112A	Synertek (4210)	60							
						SY2111A	Synertek (4210)								
						SY2101A	Synertek (4210)								
(Continued)															
256x4								(Cont'd)							
	400 nsF	CMOS	TS	22	5	HMMP1822C	Hughes	65							
						CDP1822C	RCA (1594)								
						AM9101B	AMD								
						AM9101BM	AMD								
						AM9111B	AMD								
						AM9111BM	AMD	70							
						AM9112B	AMD								
						AM9112BM	AMD								
						AM91L01B	AMD								
						AM91L01BM	AMD								
						AM91L11B	AMD								
						AM91L11BM	AMD	75							
						AM91L12B	AMD								
						AM91L12BM	AMD								
	450 nsF	CMOS	TS	22	5	S5101L-1	AMI	80							
						S5101L1	AMI								
						S6501L1	AMI								
						HM435101-1	Hitachi								
						M5L5101L-1	Mitsubishi								
						MCM51L01-45	Motorola								
						μPD5101L-1	NEC								
						SCM5101-1	SSS								
						TC5501	Toshiba								
													SY2112A-4	Synertek (4210)	85
													SY2111A-4	Synertek (4210)	
													SY2101A-4	Synertek (4210)	
							500 nsF		NMOS	TS	16	5	AM9112A	AMD	90
AM9112AM	AMD														
						AM91L12A	AMD								
						AM91L12AM	AMD								
						SY2112-1	Synertek (4210)								
				18	5	AM9101A	AMD	95							
						AM9101AM	AMD								
						AM9111A	AMD								
						AM9111AM	AMD								
						AM91L01A	AMD								
						AM91L01AM	AMD	100							
						AM91L111AM	AMD								
						AM91L11A	AMD								
						SY2111-1	Synertek (4210)								
						SY2101-1	Synertek (4210)								
	650 nsF	CMOS	TS	22	5	S5101L	AMI	105							
						S5101L-3	AMI								
						S5101L3	AMI								
						S6501L	AMI								
						S6501L-3	AMI								
						MCM5101-65	Motorola								
						MCM51L01-65	Motorola								
						μPD5101L	NEC								
						SCM5101-3	SSS								
						TC5501-1	Toshiba								
							800 nsF		CMOS	TS	22	5	S5101-8	AMI	115
S5101L8	AMI														
S6501-8	AMI														
S6501L-8	AMI														
MCM5101-80	Motorola														
MN5101	Panasonic														
LH-5101W	Sharp														
883/5101-4	SSS														
	950 nsF	NMOS	TS	18	5			M120					SGS	120	
256x8															
	40 nsF	TTL	TS	22	5			M8X350					Signetics (1627,1652)	125	
						S8X350	Signetics (527,532,1627,1652)								
256x9															
	35 nsF	TTL	TS	22	5	N82S212A	Signetics (4162)	60							
						93479	Fairchild								
	45 nsF	TTL	TS	22	5	N82S210	Signetics (Continued)								

† Military Temperature Range (-55° to 125°C)

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range

nsR—Nanoseconds at Room Temperature

OC—Open Collector

TS—Three-State

OE—Open Emitter

MEMORY-RAMs (Cont'd)

Organization	Access Time (Max)	Type	No. Output Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	No. Output Pins	Supply Voltage, V	Device	Source	Line	
<b>Static (Cont'd)</b>								1024x1	45 nsF	TTL	OC	16	5			(Cont'd)
256x9	70 nsF	TTL	TS	22	5	<b>8828212</b> † Signetics	(532)						93415C	Fairchild		
	90 nsF	TTL	TS	24	5	S82S210	† Signetics						HM2510-1	Hitachi		
512x4	1.2 μsF	CMOS	TS	24	5	LH-5102	Sharp						<b>MCM93415C</b>	<b>Motorola</b>	(4046)	
512x9	45 nsF	NMOS	TS	24	5	TMS2150-4	TI						AM93425C	AMD		
	55 nsF	NMOS	TS	24	5	<b>TMS2150-5</b>	<b>TI</b>	(4222)					93425C	Fairchild		
	70 nsF	NMOS	TS	24	5	<b>TMS2150-7</b>	<b>TI</b>	(4222)					HM2511-1	Hitachi		
	90 nsF	NMOS	TS	24	5	<b>TMS2150-9</b>	<b>TI</b>	(4222)					<b>MCM93425</b>	<b>Motorola</b>	(4046)	
1024x1	8 nsF	ECL	OE	16	-5.2	HM2112-1	Hitachi		50 nsF	TTL	OC	16	5	μPB2205	NEC	
	10 ns*	ECL	OE	16	-5.2	AM10415	AMD						SN74S314A	TI		
	10 nsF	ECL	OE	16	-4.5	100415B	Signetics						93425AM	† Fairchild		
					-5.2	F10415C	Fairchild						SN74S214A	TI		
				24	-4.5	F100415C	Fairchild						MCM2115A-55	Motorola		
					-5.2	<b>MCM10146A</b>	<b>Motorola</b>	(4046)					MCM2125A-55	Motorola		
				OE	16	HM2112	Hitachi		60 nsF	TTL	OC	16	5	AM93L415C	AMD	
	15 nsF	ECL	OE	16	-4.5	<b>AM100415A</b>	<b>AMD</b>	(3910)					93415M	† Fairchild		
					-5.2	100415A	Signetics						93L415	Fairchild		
					-5.2	<b>AM10415SA</b> † AMD	(3909)						93L415C	Fairchild		
					-5.2	<b>MCM10415</b>	<b>Motorola</b>	(4047)					<b>MCM93415M</b> † Motorola	(4046)		
	20 nsF	ECL	OE	16	-4.5	HM100415	Hitachi		65 nsF	TTL	OC	16	5	AM93415M	† AMD	
					-5.2	100415	Signetics						AM93425M	† AMD		
					-5.2	<b>AM10415A</b> † AMD	(3909)		70 ns*	TTL	OC	16	5	SN74LS315	TI	
					-5.2	AM10415SA	† AMD						SN74LS215	TI		
					-5.2	F10415AC	Fairchild		70 nsF	NMOS	OC	16	5	2115A-2	Intel	
					-5.2	MBM10415AH	Fujitsu						2115AL-2	Intel		
					-5.2	HM2110-2	Hitachi						MCM2115A-70	Motorola		
					-5.2	<b>MCM10415A</b> † Motorola	(4047)						MCM21L15A-70	Motorola		
					-5.2	DM10415A	National						2125A-2	Intel		
					-5.2	10415A	Signetics						2125AL-2	Intel		
					-5.2	2125H-1	Intel						MCM2125A-70	Motorola		
					-5.2	2125H-2	Intel						MCM21L25A-70	Motorola		
	25 nsF	ECL	OE	16	-5.2	AM10415A	† AMD						HM2148	Hitachi		
					-5.2	HM2110-1	Hitachi						AM93L415M	† AMD		
					-5.2	GXB10415	Siemens						93L415M	Fairchild		
					-5.2	2115H-2	Intel						HM2510	Hitachi		
					-5.2	2125H-2	Intel						SN74S314	TI		
	29 nsF	ECL	OE	16	-5.2	<b>MCM10146</b>	<b>Motorola</b>	(4046)					AM93L425M	† AMD		
	30 nsF	NMOS	OC	16	5	2115H-3	Intel						93L425M	† Fairchild		
					5	2125H-3	Intel						HM2511	Hitachi		
					5	AM93415AC	AMD						SN74S214	TI		
					5	AM93425AC	AMD						93L425M	† Fairchild		
					5	93415AC	Fairchild						HM2511	Hitachi		
					5	93425AC	Fairchild						SN74S214	TI		
	35 nsF	ECL	OE	16	-5.2	<b>AM10415</b>	<b>AMD</b>	(3909)					IM65X08A-1	† Intersil		
					-5.2	F10415	Fairchild						IM65X08A-1l	Intersil		
					-5.2	HM2110	Hitachi						<b>CDP1821</b> † RCA	(1594)		
					-5.2	DM10415	National						10			
					-5.2	2115H-4	Intel						18	4.5-10.5	IM65X18A-1	Intersil
					-5.2	2125H-4	Intel								IM65X18A-1M	† Intersil
					-5.2	HM2510-2	Hitachi									
	40 nsF	ECL	OE	16	-5.2	<b>AM10415M</b> † AMD	(3909)									
					-5.2	<b>MCM10546</b> † Motorola										
					-5.2	AM93415AM	† AMD									
					-5.2	AM93425AM	† AMD									
					-5.2	2115A	Intel									
					-5.2	2115AL	Intel									
					-5.2	MCM2115A-45	Motorola									
					-5.2	MCM21L15A-45	Motorola									
					-5.2	2125A	Intel									
					-5.2	2125AL	Intel									
					-5.2	MCM2125A-45	Motorola									
					-5.2	MCM21L25A-45	Motorola									
					-5.2	AM93415C	AMD									
					-5.2											

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MASTER SELECTION GUIDE

MEMORY-RAMs (Cont'd)

Organi- zation	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
<b>Static (Cont'd)</b>								(Cont'd)
256x4								(Cont'd)
	200 nsF	NMOS	TS	16	5	SY21H12-2	Synertek	
				18	5	SY21H11-2	Synertek	
				22	5	SY21H01-2	Synertek	
	220 nsF	CMOS	TS	18	5	HM6561B-2	Harris (3965)	5
						HM6561B-8	Harris (3965)	
						HM6561B-9	Harris (3965)	
						IM65X61-1	Intel	
						IM65X61-1	Intersil	
						IM65X61-1M	Harris (3965)	
				22	5	HM6551B-2	Harris (3965)	10
						HM6551B-8	Harris (3965)	
						HM6551B-9	Harris (3965)	
						IM65X51-1	Intersil	
						IM65X51-1M	Intersil	
	235 nsF	CMOS	TS	18	10	IM65X61A	Intersil	15
						IM65X61AM	Intersil	
				22	10	IM65X51A	Intersil	
						IM65X51AM	Intersil	
	240 nsF	CMOS	TS	22	5	HS6551BH	Harris (4689)	
	250 nsF	CMOS	TS	22	5	MM55101E1	RCA (1594)	20
						SCM5101-1B	SSS	
				10		CDP1822	RCA (1594)	
						TCC-244	RCA	
		NMOS	TS	16	5	AM9101D	AMD	25
						AM9111D	AMD	
						AM9112D	AMD	
						SY2112A-2	Synertek (4210)	
				18	5	µPD2111AL-2	NEC	
						SY2111A-2	Synertek (4210)	
				22	5	SY2101A-2	Synertek (4210)	30
	300 nsF	CMOS	TS	18	5	HM6561-2	Harris (3965)	
						HM6561-8	Harris (3965)	
						HM6561-9	Harris (3965)	
						IM65X61	Intersil	
						IM65X61M	Intersil	
				22	5	HM6551-2	Harris (3965)	35
						HM6551-8	Harris (3965)	
						HM6551-9	Harris (3965)	
						IM65X51	Intersil	
						IM65X51M	Intersil	
						LH-5101S	Sharp	
		NMOS	TS	16	5	AM9101C	AMD	45
						AM9101CM	AMD	
						AM9111C	AMD	
						AM9111CM	AMD	
						AM9112C	AMD	
						AM9112CM	AMD	
						AM91L01C	AMD	
						AM91L01CM	AMD	
						AM91L111C	AMD	
						AM91L111CM	AMD	
						AM91L12C	AMD	
						AM91L12CM	AMD	
	350 nsF	CMOS	TS	18	5	MM6561-5	Harris (3965)	55
						IM65X61C	Intersil	
				22	5	MM6551-5	Harris (3965)	
						IM65X51C	Intersil	
						MM55101BL3	RCA (1594)	
						MM55101EL3	RCA (1594)	
						SCM5101-1A	SSS	
		NMOS	TS	16	5	SY2112A	Synertek (4210)	60
				18	5	SY2111A	Synertek (4210)	
				22	5	SY2101A	Synertek (4210)	
(Continued)								(Continued)

Organi- zation	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
	400 nsF	CMOS	TS	22	5	HMMP1822C	Hughes	(Cont'd)	
						CDP1822C	RCA	65	
		NMOS	TS	16	5	AM9101B	AMD	70	
						AM9101BM	AMD		
						AM9111B	AMD		
						AM9111BM	AMD		
						AM9112B	AMD		
						AM9112BM	AMD		
						AM91L01B	AMD		
						AM91L01BM	AMD		
						AM91L11B	AMD		
						AM91L11BM	AMD		
						AM91L12B	AMD		
						AM91L12BM	AMD		
	450 nsF	CMOS	TS	22	5	S5101L-1	AMI	80	
						S5101L1	AMI		
						S6501L1	AMI		
						HM435101-1	Hitachi		
						M5L5101L-1	Mitsubishi		
						MCM51L01-45	Motorola		
						µPD5101L-1	NEC		
						SCM5101-1	SSS		
						TC5501	Toshiba	85	
		NMOS	TS	16	5	SY2112A-4	Synertek (4210)		
				18	5	SY2111A-4	Synertek (4210)		
				22	5	SY2101A-4	Synertek (4210)		
	500 nsF	NMOS	TS	16	5	AM9112A	AMD	90	
						AM9112AM	AMD		
						AM91L12A	AMD		
						AM91L12AM	AMD		
						SY2112-1	Synertek (4210)		
				18	5	AM9101A	AMD	95	
						AM9101AM	AMD		
						AM9111A	AMD		
						AM9111AM	AMD		
						AM91L01A	AMD		
						AM91L01AM	AMD	100	
						AM91L111AM	AMD		
						AM91L111A	AMD		
						SY2111-1	Synertek (4210)		
				22	5	SY2101-1	Synertek (4210)		
	650 nsF	CMOS	TS	22	5	S5101L	AMI	105	
						S5101L-3	AMI		
						S5101L3	AMI		
						S6501L	AMI		
						S6501L-3	AMI		
						MCM5101-65	Motorola		
						MCM51L01-65	Motorola		
						µPD5101L	NEC		
						SCM5101-3	SSS		
						TC5501-1	Toshiba		
	800 nsF	CMOS	TS	22	5	S5101-8	AMI	115	
						S5101L8	AMI		
						S6501-8	AMI		
						S6501L-8	AMI		
						MCM5101-80	Motorola		
						MN5101	Panasonic		
						LH-5101W	Sharp		
						883/5101-4	SSS	120	
	950 nsF	NMOS	TS	18	5	M120	SGS		
	256x8	40 nsF	TTL	TS	22	5	88X350	Signetics (1627,1652)	
						88X350	Signetics (527,532,1627,1652)	125	
	256x9	35 nsF	TTL	TS	22	5	882S212A	Signetics (4162)	
		45 nsF	TTL	TS	22	5	93479	Fairchild	
		60 nsF	TTL	TS	24	5	882S210	Signetics	(Continued)

† Military Temperature Range (-55° to 125°C) ns\*—Nanoseconds Typical nsF—Nanoseconds over Full Temperature Range nsR—Nanoseconds at Room Temperature  
OC—Open Collector TS—Three-State OE—Open Emitter

# IC MASTER

## MEMORY-RAMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line		
<b>Static (Cont'd)</b>									<b>(Cont'd)</b>										
256x9	70 nsF	TTL	TS	22	5	S82S212	† Signetics (532)			45 nsF	TTL	OC	16	5	93415C	Fairchild			
	90 nsF	TTL	TS	24	5	S82S210	† Signetics								HM2510-1	Hitachi			
512x4	1.2 μsF	CMOS	TS	24	5	LH-5102	Sharp								MCM93415C	Motorola (4046)	65		
512x9	45 nsF	NMOS	TS	24	5	TM52150-4	TI								AM93425C	AMD			
	55 nsF	NMOS	TS	24	5	TMS2150-5	TI (4222)	5							93425C	Fairchild			
	70 nsF	NMOS	TS	24	5	TMS2150-7	TI (4222)								HM2511-1	Hitachi			
	90 nsF	NMOS	TS	24	5	TMS2150-9	TI (4222)								MCM93425	Motorola (4046)	70		
1024x1	8 nsF	ECL	OE	16	-5.2	HM2112-1	Hitachi			50 nsF	TTL	OC	16	5	μPB2205	NEC			
	10 ns*	ECL	OE	16	-5.2	AM10415	AMD								SN74S314A	TI			
	10 nsF	ECL	OE	16	-4.5	100415B	Signetics	10							93425AM	† Fairchild			
					-5.2	F10415C	Fairchild								SN74S214A	TI			
				24	-4.5	F100415C	Fairchild								MCM2115A-55	Motorola			
					-5.2	MCM10146A	Motorola (4046)								MCM2125A-55	Motorola	75		
				OE	16	HM2112	Hitachi	15							AM93L415C	AMD			
	15 nsF	ECL	OE	16	-4.5	AM100415A	AMD (3910)								93415M	† Fairchild			
					-5.2	100415A	Signetics								93L415	Fairchild			
					-5.2	AM10415SA	† AMD (3909)								93L415C	Fairchild			
					-5.2	MCM10415	Motorola (4047)								MCM93415M	† Motorola (4046)	80		
	20 nsF	ECL	OE	16	-4.5	HM100415	Hitachi	20							93425M	† Fairchild			
					-5.2	100415	Signetics								93L425C	Fairchild			
					-5.2	AM10415A	† AMD (3909)								AM93415M	† AMD			
					-5.2	AM10415SA	† AMD (3909)								AM93425M	† AMD			
					-5.2	F10415AC	Fairchild								SN74LS315	TI			
					-5.2	MBM10415AH	Fujitsu								SN74LS215	TI	85		
					-5.2	HM2110-2	Hitachi								2115A-2	Intel			
					-5.2	MCM10415A	† Motorola (4047)								2115AL-2	Intel			
					-5.2	DM10415A	National								MCM2115A-70	Motorola			
					-5.2	10415A	Signetics								MCM21L15A-70	Motorola	90		
					-5.2	2125H-1	Intel								2125A-2	Intel			
					-5.2	AM10415A	† AMD (3909)								2125AL-2	Intel			
					-5.2	HM2110-1	Hitachi								MCM2125A-70	Motorola			
					-5.2	GXB10415	Siemens								MCM21L25A-70	Motorola			
					-5.2	2115H-2	Intel								HM2148	Hitachi	95		
					-5.2	2125H-2	Intel								AM93L415M	† AMD			
					-5.2	MCM10146	Motorola (4046)								93L415M	Fairchild			
					-5.2	2115H-3	Intel								HM2510	Hitachi			
					-5.2	2125H-3	Intel								SN74S314	TI			
					-5.2	AM93415AC	AMD								AM93L425M	† AMD	100		
					-5.2	AM93425AC	AMD								93L425M	† Fairchild			
					-5.2	93415AC	Fairchild								HM2511	Hitachi			
					-5.2	93425AC	Fairchild								SN74S214	TI			
					-5.2	AM10415	AMD (3909)								SN74LS314	TI			
					-5.2	F10415	Fairchild								SN74LS214	TI			
					-5.2	HM2110	Hitachi								IM65X08A-1	† Intersil			
					-5.2	DM10415	National								IM65X08A-1I	Intersil			
					-5.2	2115H-4	Intel								10	CDP1821	† RCA (1594)		
					-5.2	2125H-4	Intel								18	4.5-10.5	IM65X18A-1	Intersil	
					-5.2	HM2510-2	Hitachi								18	4.5-10.5	IM65X18A-1M	† Intersil	110
					-5.2	AM10415M	† AMD (3909)										NMC85008B-9	National	
					-5.2	MCM10546	† Motorola										HM6508B-2	† Harris (3965)	
					-5.2	AM93415AM	† AMD										HM6508B-8	† Harris (3965)	
					-5.2	AM93425AM	† AMD										HM6508B-9	Harris (3965)	
					-5.2	2115A	Intel										IM5X08-1	Intersil	
					-5.2	2115AL	Intel										IM65X08-1M	† Intersil	115
					-5.2	MCM2115A-45	Motorola										IM6518B-2	† Harris (3965)	
					-5.2	MCM21L15A-45	Motorola										HM6518B-8	† Harris (3965)	
					-5.2	2125A	Intel										HM6518B-9	Harris (3965)	
					-5.2	2125AL	Intel										IM65X18-1	Intersil	
					-5.2	MCM2125A-45	Motorola										IM65X18-1M	† Intersil	120
					-5.2	MCM21L25A-45	Motorola										IM65X08A	Intersil	
					-5.2	AM93415C	AMD										IM65X08AM	† Intersil	
					-5.2												IM65X18A	Intersil	
					-5.2												IM65X18AM	† Intersil	
					-5.2												HM6508-2	† Harris (3965)	125

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MEMORY-RAMs (Cont'd)

Table with columns for Organization, Access Time (Max), Type, Output, No. Pins, Supply Voltage (V), Device, Source, and Line. It is divided into two main sections: Static (left) and 1024x4 (right). The Static section lists various RAM models like HM6508-8, IM65X08, etc. The 1024x4 section lists models like MBM10474-10, AM100474A, etc. The table includes line numbers on the right margin and is marked as '(Continued)' at the bottom.

† Military Temperature Range (-55° to 125°C) OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range TS—Three-State

nsR—Nanoseconds at Room Temperature OE—Open Emitter



MEMORY-RAMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
<b>Static (Cont'd)</b>									<b>(Cont'd)</b>									
1024x4	300 nsF	NMOS	TS	22	5	AM9130C AM9130CM †AMD AM91L30C AM91L30CM †AMD	AMD	(Cont'd)	1024x8	90 nsF	NMOS	TS	24	5	SY2159B-1 SY2159L-1	Syaertek Syaertek	(4210) (4210)	65
										100 nsF	Dual Port	—	48	5	SY21D1 SYM21D1 †	Synertek Synertek	(4210)	70
											NMOS	—	48	5	SY2130 SY2131	Syaertek Syaertek	(4210) (4210)	70
	350 nsF	CMOS	TS	18	5	HM6514-5 NMC6514-5	Harris National	(3966)		120 nsF	NMOS	TS	24	5	MK4118A-1 MKB4801A-81 †	Mostek Mostek	(4210)	75
	400 nsF	NMOS	TS	22	5	AM9130B AM9130BM †AMD AM91L30B AM91L30BM †AMD	AMD								SY2158A-2 SY2158B-2 SY2158L-2 SY2159A-2 SY2159B-2 SY2159L-2	Syaertek Syaertek Syaertek Syaertek Syaertek Syaertek	(4210) (4210) (4210) (4210) (4210) (4210)	75
	450 nsF	CMOS	TS	18	5	HM4334-4 M58981-45 μPD444/6514	Hitachi Mitsubishi NEC			150 nsF	NMOS	TS	22	5	μPD421-5 MK4118A-2 SY2158A-3 SY2158B-3 SY2158L-3 SY2159A-3 SY2159B-3 SY2159L-3	NEC Mostek Syaertek Syaertek Syaertek Syaertek Syaertek Syaertek	(4210) (4210) (4210) (4210) (4210) (4210) (4210) (4210)	80
																		85
		NMOS	TS	18	5	AM9114B AM9114BM †AMD AM9124B AM9124BM †AMD AM91L14B AM91L14BM †AMD AM91L24B AM91L24BM †AMD	AMD											90
						2114 2114L 2114M †	Fairchild											95
						HM472114-4 MSL2114L MCM2114-45 MCM2114-45 MM2114 MM2114L EA2114L μPD2114L SY2114 SY2114L SYM2114 † TMS2114-45 TMS2114L-45 TMM314A TMM314AL	Hitachi Mitsubishi Motorola Motorola National National NEC NEC Syaertek Syaertek Syaertek TI TI Toshiba Toshiba	(4046) (4046)										100
																		105
	500 nsF	NMOS	TS	22	5	AM9130A AM9130AM †AMD AM91L30A AM91L30AM †AMD	AMD			250 nsF	NMOS	TS	22	5	μPD421-2 MK4118A-4	NEC Mostek	(4210) (4210)	90
	550 nsF	CMOS	TS	20	5	TC5047A-1	Toshiba											95
	650 nsF	CMOS	TS	20	5	μPD445L-1 TC5514-1	NEC Toshiba											100
	800 nsF	CMOS	TS	20	5	TC5047A-2 TC5514-2	Toshiba			300 nsF	NMOS	TS	22	5	μPD421-1 R8104-3	NEC Rockwell	(4210) (4210)	90
	1000 nsF	CMOS	TS	18	1-1.7	MP2114CLP1 MP6514LP1	MicroPwr											105
						MP2114CLP3	MicroPwr			400 nsF	NMOS	TS	22	5	R8114-3	Rockwell		100
1024x8	55 nsF	NMOS	TS	24	5	MK4801A-55	Mostek		2048x1	300 nsF	CMOS	TS	18	5	NMC6503-9 † NMC6503-5	National National	(535)	35
	70 nsF	NMOS	TS	24	5	MK4801A-70	Mostek			350 nsF	CMOS	TS	18	5				30
	90 nsF	NMOS	TS	24	5	MKB4801A-870 † MK4801A-90 MKB4801A-990 †	Mostek Mostek Mostek											40
						SY2158A-1 SY2158B-1 SY2158L-1 SY2159A-1	Syaertek Syaertek Syaertek Syaertek	(4210) (4210) (4210) (4210)										45
																		50
																		55
																		60
																		65
																		70
																		75
																		80
																		85
																		90
																		95
																		100
																		105
																		110
																		115
																		120
																		125

† Military Temperature Range (-55° to 125°C) ns—Nanoseconds Typical nsF—Nanoseconds over Full Temperature Range nsR—Nanoseconds at Room Temperature  
OC—Open Collector TS—Three-State OE—Open Emitter





# MASTER SELECTION GUIDE

## MEMORY-RAMs (Cont'd)

Organi- zation	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organi- zation	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line			
<b>Static</b>									<b>(Cont'd)</b>											
2048x8									4096x1	35 nsF	NMOS	TS	18	5	2147H-1		Intel	(Cont'd)		65
															μPD2147A-35		NEC			
															SY2147H-1		Synertek	(4210)		
										40 nsF	ECL	OE	18	-4.5	AM10470M		↑AMD	(3913)		
										45 nsF	CMOS	TS	18	5	HM6147-45		Hitachi	70		
															AM2147-45		AMD			
															AM2147-45M		↑AMD			
															AM21L47-45		AMD			
															HM4847-2		Hitachi			
															2147H-2		Intel			
															M2147H-2		↑Intel			
															NM2148H-2		National			
															μPD2147A-45		NEC	75		
															SY2147H-2		Synertek		(4210)	
															SYM2147H-2		↑Synertek		(535)	
										55 nsF	CMOS	TS	18	5	HM6147-3		Hitachi	80		
															HM6147-55		Hitachi			
															MCM6147-55		Motorola	(4046)		
															MCM61L47-55		Motorola			
															AM2147-55		AMD	85		
															AM2147-55M		↑AMD			
															AM21L47-55		AMD			
															HM4847-3		Hitachi			
															2147-3		Intel			
															2147A-3		Intel			
															2147AL-3		Intel			
															2147H-3		Intel			
															2147HL-3		Intel	90		
															M2147H-3		↑Intel			
															ITT2147-55		ITT			
															MCM2147-55		Motorola	95		
															MM2147-3		National			
															MM2148H-3		National			
															MM2148H-3L		National			
															NMC2147H-3		National			
															NMC2147H-4		National			
															μPD2147-3		NEC			
															SY2147-3		Synertek	(4210)		
															SY2147H-3		Synertek	(4210)		
															SY2147HL-3		Synertek	(4210)		
															SYM2147-3		↑Synertek	(535)		
															SYM2147-6		↑Synertek	(535)		
															SYM2147H-3		Synertek	(535)		
															TMM315-1		Toshiba	105		
															UM2147-1		Universal			
										70 nsF	CMOS	TS	18	5	HM6147		Hitachi	110		
															HM6147L		Hitachi			
															MCM6147-70		Motorola	(4046)		
															MCM61L47-70		Motorola			
															AM2147-70		AMD	115		
															AM2147-70M		↑AMD			
															AM21L47-70		AMD			
															HM4847		Hitachi			
															2147A		Intel			
															2147AL		Intel			
															2147H		Intel			
															2147HL		Intel			
															M2147H		↑Intel	120		
															2147		Intersil			
															ITT2147-70		ITT			
															MCM2147-70		Motorola	125		
															MM2147		National			
															MM2147L		National			
															MM2148H		National			
															MM2148H-L		National			
															μPD2147-2		NEC			
															SY2147H		Synertek	(4210)		
(Continued)									(Continued)											

† Military Temperature Range (-55° to 125°C)    ns—Nanoseconds Typical    nsF—Nanoseconds over Full Temperature Range    nsR—Nanoseconds at Room Temperature  
 OC—Open Collector    TS—Three-State    OE—Open Emitter





MEMORY-RAMs (Cont'd)

Organization	Access Time (Max)	Type	No. Output Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	No. Output Pins	Supply Voltage, V	Device	Source	Line
<b>Static (Cont'd)</b>															
8192x8	400 nsF	CMOS Module	40	5				16384x1	85 nsF	CMOS	TS	22	5		(Cont'd)
		TS	40	5	NM5-6564-2 † Harris	(3968)				NMOS	TS	20	5	IDT7M464-85 IDT	(4004)
					NM5-6564-9 Harris	(3968)								IDT7M464-85C † IDT	(4004)
	450 nsF	CMOS Module	40	5										MBM8167A-85W	
		TS	40	5	NM5-6564-5 Harris	(3968)								† Fujitsu	
16384x1	15 nsF	ECL	20	-4.5	MB81256-10	Fujitsu	5		100 nsF	CMOS	TS	20	5	HM6167-8	Hitachi
				-5.2	MBM100480-15	Fujitsu						22	5	HM6167L-8	Hitachi
					MBM10480	Fujitsu						24	5	IDT6167S100	IDT (3988)
	25 nsF	ECL	OE	20	MBM100480	Fujitsu								IDT7M464-100C † IDT	(4004)
				-4.5	HM100480	Hitachi								IDT6167L100 † IDT	(3988)
				-5.2	MBM10480	Fujitsu								IDT6167L100	IDT
	35 nsF	ECL	OE	20	F100480	Fairchild	10							IMS1400-10L	Immos (3977)
				-5.2	F10480	Fairchild								AM9016F	AMD (3918)
		NMOS	TS	20	AM2167-35	AMD (3915)		16384x4	350 nsF	CMOS	TS	40	5	HM5-6564-2 † Harris	(3968)
					IMS1400-35	Immos (3977)								HM5-6564-9 Harris	(3968)
					MCM2167H-35	Motorola (4046,4053)			450 nsF	CMOS	TS	40	5	HM5-6564-5 Harris	
45 nsF	CMOS	TS	20	5	HM6167-45	Hitachi	15	16384x16	150 nsF	CMOS	TS	48	5	HM92560-8 † Harris	(3968)
					IDT6167L45	IDT (3988)								HM92560-2 † Harris	(3968)
					IDT6167S45	IDT (3988)								HM92560-9 Harris	(3968)
		NMOS	TS	20	AM2167-45	AMD (3915)			250 nsF	CMOS Module	TS	48	5	HM92560-5 Harris	(3968)
					AM2167-45M † AMD	(3915)								HM92570-2 † Harris	(3968)
					MB8167A-45	Fujitsu	20							HM92570-8 † Harris	(3968)
					IMS1400-45	Immos (3977)								HM92570-9 Harris	(3968)
					ITT2167-45	ITT								HM92560-2 † Harris	(3968)
					MCM2167H-45	Motorola (4046,4053)								HM92560-8 † Harris	(3968)
55 nsF	CMOS	TS	20	5	HM6167-55	Hitachi	25							HM92560-9 Harris	(3968)
					IDT6167L55	IDT (3988)								HM92570-2 † Harris	(3968)
					IDT7M464-55	IDT (3988)								HM92570-8 † Harris	(3968)
					IDT6167S55 † IDT	(3988)								HM92570-9 Harris	(3968)
					IDT6167S55	IDT								HM92570-9 Harris	(3968)
		NMOS	TS	20	AM2167-55	AMD (3915)								HM92570-5 † Harris	(3968)
					AM2167-55M † AMD	(3915)		32768x8	150 nsF	CMOS	TS	48	5	HM92560-8 † Harris	(3968)
					MB8167A-55	Fujitsu								HM92560-9 Harris	(3968)
					IMS1400-55	Immos (3977)								HM92570-2 † Harris	(3968)
					IMS1400-55M † Immos	(501)								HM92560-5 Harris	(3968)
					ITT2167-55	ITT								HM92570-8 † Harris	(3968)
					μPD2167	NEC (4086)								HM92570-9 Harris	(3968)
					μPD2167-3	NEC (4086)								HM92570-2 † Harris	(3968)
					SY2167-3	Synertek (4210)								HM92570-8 † Harris	(3968)
65 nsF	CMOS	TS	22	5	IDT7M464-65	IDT (4004)	40							HM92570-9 Harris	(3968)
					IDT7M464-65C † IDT	(4004)								HM92570-5 † Harris	(3968)
70 nsF	CMOS	TS	20	5	HM6167	Hitachi	45	65536x1	70 nsF	CMOS	TS	22	5	IDT7M164-70	IDT (4008)
					IDT6167L70	IDT (3988)								IDT7M164-100	IDT (4008)
					IDT6167S70	IDT (3988)								IDT7M164-100C † IDT	(4008)
		NMOS	TS	20	AM2167-70	AMD (3915)								AM9064-10	AMD (3919)
					AM2167-70M † AMD	(3915)								IDT7M164-120	IDT (4008)
					MB8167A-70	Fujitsu								IDT7M164-120C † IDT	(4008)
					MB8167A-70W	Fujitsu								AM9064-12	AMD (3919)
					IMS1400-70L	Immos (501)								IDT7M164-150	IDT (4008)
					IMS1400-70M † Immos	(3977)								IDT7M164-150C † IDT	(4008)
					IMS1400L	Immos (3977)								AM9064-15 † AMD	(3919)
					ITT2167-70	ITT								ITT4164-15	ITT
					μPD2167-2	NEC (4086)								ITT4164-20	ITT
					SY2167	Synertek (4210)									
					SYM2167	† Synertek (535)									
85 nsF	CMOS	TS	20	5	HM6167-6	Hitachi	55								
					HM6167L-6	Hitachi									
					IDT6167L85 † IDT	(3988)									
					IDT6167L85	IDT									
					IDT6167S85 † IDT	(3988)									
					IDT6167S85	IDT	60								

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
32x8	45 nsR	TTL	OC	16	5	SN7488A	TI	
256x4	1.8 $\mu$ sR	CMOS	—	16	3-15	HD14524B MCM14524A † Motorola (4048) MCM14524C Motorola (4048)	Hitachi	
	60 nsF	TTL	OC	16	5	SN74187	TI	5
256x8	70 nsF	TTL	OC TS	20 5	5	SN74S271 SN74S371	TI	
512x4	70 nsF	TTL	OC TS	16 5	5	SN74S270 SN74S370	TI	
512x8	70 nsF 90 nsF 500 nsF	NMOS	TS	24 24 24	5 5 10	82HM141C 82HM141M HCMP1831 HCMP1832 HMMP1831 HMMP1832 CDP1831 CDP1832	Intersil Intersil Hughes Hughes Hughes Hughes RCA (1593) RCA (1593)	10 15
	700 nsF 1000 nsF	PMOS CMOS	TS	24 24	—12,5 5	3514 HCMP1831C HCMP1832C HMMP1831C HMMP1832C CDP1831C CDP1832C	Fairchild Hughes Hughes Hughes Hughes RCA (1593) RCA (1593)	20
1024x4	60 nsF 80 nsF	HMOS	TS	24 24	5 5	82HM137C 82HM137M	Intersil Intersil	25
1024x8	35 nsF 45 nsF	NMOS	TS	24 22	5 5	SY3308R MB7130E MB7130H	Synertek (4211) Fujitsu Fujitsu	
	55 nsF	TTL	OC TS	24 24	5 5	6280-2 6281-2	MMI MMI	30
	70 nsF	HMOS NMOS	TS	24 24	5 5	82HM181C SY3308	Intersil Synertek (4211)	
	75 nsF	TTL	OC TS	24 24	5 5	5280-2 5281-2	† MMI † MMI	35
	80 nsF	TTL	OC TS	24 24	5 5	6280-1 6282-1 6281-1 6283-1	MMI MMI MMI MMI	40
	90 nsF	HMOS TTL	TS OC TS	24 24 24	5 5 5	82HM181M DM75S29 DM75S28	† Intersil † National † National	
	175 nsF	TTL	OC TS	24 24	5 5	5280-1 5282-1 5281-1 5283-1	† MMI † MMI † MMI † MMI	45
	250 nsF	NMOS	TS	24	5	F68B308 MCM68B308 MCM68B30A	Fairchild Motorola Motorola	50
	350 ns *	CMOS	TS	24	10	HCMP1833 HCMP1834 HMMP1833 HMMP1834 CDP1833 CDP1834	Hughes Hughes Hughes Hughes RCA (1593) RCA (1593)	55
	350 nsF	NMOS	TS	24	5	MCM68A30A $\mu$ PD2308A-35 SNJ2708-35 † TI TMS2708-35	Motorola NEC (4096) (4234) TI	60
	450 nsF	NMOS	TS	24	5	MCM68A306 MCM68A306P7 $\mu$ IPD2308A-45	Motorola Motorola NEC (4096)	(Continued)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
1024x8	450 nsF	NMOS	TS	24	5	SMJ2708-45 † TI SMJ27108-45 † TI TMS2708-45 TMS27108-45	TI TI TI TI	(Cont'd) 65
					$\pm$ 5,12	AM8308 $\mu$ PD2308A	AMD NEC (4096)	
	500 nsF	NMOS	TS	24	5	HN46830	Hitachi	
	850 ns *	CMOS	TS	24	5	CDP1833C CDP1834C	† RCA (1593) † RCA (1593)	70
		NMOS	TS	24	5	HCMP1833C HCMP1834C HMMP1833C HMMP1834C	Hughes Hughes Hughes Hughes	75
1024x9	100 nsF	TTL	OC TS	24 24	5 5	6260-1 6261-1	MMI MMI	
	150 nsF	TTL	OC TS	24 24	5 5	5260-1 5261-1	† MMI † MMI	80
1024x10	100 nsF	TTL	OC TS	24 24	5 5	6255-1 6256-1	MMI MMI	
	150 nsF	TTL	OC TS	24 24	5 5	5255-1 5256-1	† MMI † MMI	85
1024x12	200 nsR 220 nsR 510 nsF 560 nsF 640 nsF	CMOS	TS	18 18 18 18 18	10 5 5 5 5	IM6312AI IM6312AM IM6312-1I IM6312-1M IM6312M	Intersil Intersil Intersil Intersil Intersil	90
1152x9	125 nsF	TTL	OC	24	5	6290 6291	MMI MMI	
2048x4	60 nsF	NMOS	TS	24	5	72HM185C 82HM185M	Intersil Intersil	95
2048x8	— 35 nsF 45 nsF 55 nsF 80 nsF	NMOS	— TS TS TS	16 24 22 22 24	5 5 5 5 5	SPR-16 SY3316R MB7136H MB7136E MB7138E-W	GI Synertek (4211) Fujitsu Fujitsu Fujitsu	100
	110 nsF	NMOS	TS	24	5	82HM191C SYM3316 SYM3316A SY2316A SY3316 SY3316A	Intersil MMI MMI Synertek (4211) Synertek (4211) Synertek (4211)	105
	100 nsF	NMOS	TS	24	5	82HM191M SYM3316 SYM3316A	† Intersil † Synertek (535) † Synertek (535)	110
	110 nsF	NMOS TTL	TS OC TS	24 24 24	5 5 5	S68116 6275-1 6276-1	AMI MMI MMI	115
	120 nsF	TTL	OC TS	24 24	5 5	5275-1 5276-1	† MMI † MMI	
	150 ns *	I <sup>2</sup> L	OC/TTL	24	5	SBP9818C SBP9818M	TI TI	120
	200 nsF	NMOS	TS	24	5	SY2316B-2	Synertek (4211)	
	250 nsF	CMOS	TS	24	5	Z3SC16A F3531B-25 2616A-25	Supertex Fairchild (1464) Signetics	
	300 nsF	CMOS	TS	24	5	MD23SC16 CDP1835C	Mitel † RCA (1593)	
		HMOS	TS	24	5	IM7332C	Intersil	
		NMOS	TS	24	5	F3531B-30 G2316B-3 2616-30 SY2316B-3	Fairchild (1464) GTEMicro Signetics Synertek (4211)	125

† Military Temperature Range (-55° to 125°C)  
OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range  
TS—Three-State

nsR—Nanoseconds at Room Temperature  
OE—Open Emitter

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## MEMORY-ROMs (Cont'd)

Organiza- tion	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organiza- tion	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
2048x8	350 ns*	CMOS	TS	24	5	IM63161 IM6316M	Intersil † Intersil	(Cont'd)	4096x8	200 nsF	NMOS	TS	24	5	SY2332-2 SY2333-2 SYM2332-3 SYM2333-3	Syaertek Syaertek † Syaertek † Syaertek	(4211) (4211) (535) (535)	
		PL	OC/TTL	24	5	SBP8316C SBP8316M	TI † TI			250 nsF	CMOS	TS	24	5	23SC32A 23SC33A	Supertex Supertex	70	
	350 nsF	CMOS	TS	24	5	23MC16A 23SC16A-3	† Supertex Supertex	5			NMOS	TS	24	5	2332-25 2632A-25 UM2333-25	Signetics Signetics Universal		
		NMOS	TS	24	5	AM8316EC AM9218C F3516-35 F68316 R03-9316C MK34000-3 MCM68A316A MCM68A316E MCM68A316EP91	AMD AMD Fairchild Fairchild (1464) GI Mostek Motorola (4048) Motorola (4048) Motorola	10		300 nsF	CMOS	TS	24	5	883/23C32M 883/23C33M SCM23C32 SCM23C32M SCM23C33 SCM23C33M	† SSS † SSS SSS † SSS SSS † SSS	75	
						μPD2316E-35 μPD2316E-5 2616-35 SMJ2516-35 TMS2516-35	NEC (4098) NEC (4098) Signetics † TI (4232) TI (543,4232)	15			NMOS	TS	24	5	AM9232C AM9233C F3532-30 F3533-30 IM7332	AMD AMD Fairchild (1465) Fairchild (1465) Intersil	85	
	430 nsF	CMOS	—	18	5	MCM65516-43	Motorola (4048)	20							μPD2332A-30	NEC (4100)		
	450 nsF	NMOS	TS	24	5	AM8316EB AM9217BC AM9217BM AM9218BC AM9218BM S68A316 S68B316 F3516-45 R03-9316B G2316B-4 MK34000-84 MM52116 μPD2316E	AMD AMD † AMD AMD † AMD AMI AMI Fairchild GI GTEMicro † Mostek National NEC (4098)	25						μPD2332B-30 R2332-3 2332-30 2632A-30 2632AE-30 SY2332-3 SY2333-3 SMJ2532-30 TMS2532-30 TMS2732A-30 TMS4732-30 VT2332A VT2333A	NEC (4100) Rockwell Signetics Signetics † Signetics Syaertek (4211) Syaertek (4211) † TI (4231) TI (543,4231) TI (4230) TI (4226) VTI (4249) VTI (4249)	90		
						μPD2316E-45 R2316B R2316E M2316 2616-45 SY2316B SMJ2516-45 TMS2516-45 TMM331A TMM334 TSU2316E	NEC (4098) Rockwell Rockwell SGS Signetics Syaertek (4211) † TI (4232) TI (543,4232) Toshiba Toshiba Toshiba	30		350 nsF	CMOS	TS	24	4-6.5	CDM5332 CDM5333	RCA RCA	(1593) (1593)	100
						23MC32A 23MC33A 23SC32A-3 23SC33A-3	† Supertex † Supertex Supertex Supertex	35							23MC32A 23MC33A 23SC32A-3 23SC33A-3	† Supertex † Supertex Supertex Supertex	105	
						2616-45 SY2316B SMJ2516-45 TMS2516-45 TMM331A TMM334 TSU2316E	Signetics Syaertek (4211) † TI (4232) TI (543,4232) Toshiba Toshiba Toshiba	40			NMOS	TS	24	5	S2333 S68A332 F3532-35 F3533-35 F68A332 G2332-3 MCM68A332 MCM68A332P2 μPD2332A-1	AMI (3932) AMI Fairchild (1465) Fairchild (1465) Fairchild GTEMicro Motorola (4048) Motorola NEC (4100)	110	
	550 nsF	CMOS	—	18	5	MCM65516-55	Motorola (4048)	45							μPD2332A-35 μPD2332B-1	NEC (4100) NEC (4100)	115	
		TS		18	4.5-5.5	CDP65516	RCA (1593)								IPD2332B-35 2632A-35 2632AM-35	NEC (4100) Signetics † Signetics	120	
		NMOS	TS	24	5	CM1600	Supertex								SMJ2532-35 TMS2532-35 TMS2732A-35 TMS4732-35 TMM2332	† TI (4231) TI (543,4231) TI (4230) TI (4226) Toshiba		
	800 nsF	CMOS	TS	24	5	AM9217AC AM9217AM	AMD † AMD	50		450 nsF	CMOS	TS	24	5	TC5332 TC5333 TC5334 TC5335	Toshiba Toshiba Toshiba Toshiba	125	
2048x10	500 nsF	NMOS	TS	40	5	CM1600-2 MM1600	Supertex † Supertex				NMOS	TS	24	5	AM9232BC	AMD	(Continued)	
						R03-9502	GI											
4096x4	45 nsF	TTL	TS	20	5	MB7134H MB7134E	Fujitsu Fujitsu	55										
	55 nsF	TTL	TS	20	5	MB7142E-W S68132	† Fujitsu AMI											
4096x8	65 nsF	TTL	TS	24	5	SCM23C32-3 SCM23C33-3	SSS SSS	60										
	115 nsF	NMOS	TS	24	5	SCM23C32-4 SCM23C33-4	SSS SSS											
	150 nsF	CMOS	TS	24	5	R09432D R09433D 2332-20 2632A-20	GI GI Signetics Signetics	65										
	200 nsF	CMOS	TS	24	5													

\* Military Temperature Range (-55° to 125°C)

\* Typical Value

† Bold face indicates additional data is provided on the page noted.

Master Selection Guide

MEMORY

MEMORY-ROMs (Cont'd)

Organi- zation	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organi- zation	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	
(Cont'd)									(Cont'd)									
4096x8	450 nsF	NMOS	TS	24	5	AM9232BM †	AMD		8192x8	200 nsF	NMOS	TS	28	5	µPD2364E	NEC	(4102,4104)	
						AM9233BC	AMD		IPD2364E-20	NEC	(4102)	65						
						AM9233BM †	AMD		R2364B2	Rockwell								
						S68332	AMI		2364-20	Signetics	(4196)							
						F3532-45	Fairchild (1465)		5	SY6365	Synortek	(4211)						
						F3533-45	Fairchild (1465)		10	TMS2764H-20	TI	(4228)						
						R09433C	GI		70	TMM2365	Toshiba							
						R09432B	GI		VT2365B	VTI	(4251)							
						R09433B	GI		10	VT2366B	VTI	(4251)						
						G2332-4	GTEMicro		250 nsF	CMOS	TS	24	4-6					
						MM52132	National		5	CDM5364	RCA	(1593)						
						µPD2332	NEC		(4100)	MP2364-25	MicroPwr							
						µPD2332A	NEC		(4100)	MP2365-25	MicroPwr	75						
						µPD2332A-45	NEC		(4100)	TC5364	Toshiba							
						µPD2332B	NEC		(4100)	TC5365	Toshiba							
						µPD2332B-45	NEC		(4100)	TC5366	Toshiba							
						R2332	Rockwell		28	4-6	CDM5365	RCA	80					
						2332-45	Signetics			NMOS	TS	24	5					
						2632A-45	Signetics		20	AM9264B	AMD	(3922)						
						2632AM-45 †	Signetics			S68B364	AMI	(3933)						
						ROM4732	SMC		25	F3564-25	Fairchild (1465)							
						SY2332	Synortek (4211)		25	MK36000-4	Mostek							
						SY2333	Synortek (4211)			MKB36000-83 †	Mostek	85						
						SYM2332 †	Synortek (535)			MKB37000-84 †	Mostek							
						SYM2333 †	Synortek (535)			MCM68364-25	Motorola	(4048,4054)						
						\$MJ2532-45 †	TI (4231)		25	MCM68365-25	Motorola	(4048,4054)						
						TMS2532-45	TI (543,4231)			MCM68366-25	Motorola	(4048,4054)						
						TMS2732A-45	TI (4230)			MCM68B364	Motorola	90						
						TMS4732	TI (4226)		30	R2364A25	Rockwell							
						TMS4732-45	TI (4226)			M36000	SGS							
						TMM333	Toshiba			2664-25	Signetics							
						TSU333-2	Toshiba			2664A-25	Signetics							
						VT2332	VTI (4249)			ROM36000	SMC	95						
						VT2333	VTI (4249)			SY2364A	Synortek (4211)							
									35	VT2364B	VTI (4250)							
	500 nsF	NMOS	TS	24	5	F68332	Fairchild (1465)		28	5	AM9265D	AMD	(3923)					
	500 µsF	CMOS	TS	24	4-6.5	CDP1837C	RCA (1593)		S2364B	AMI	(3934)							
	800 nsF	CMOS	TS	24	5	CM3200-2	Supertex		F3565-25	Fairchild (1465)	100							
	1500 nsF	CMOS	TS	24	5	MM3200	† Supertex		MK37000-4	Mostek								
						CM3200-3	Supertex		MK37000-5	Mostek								
4096x10	300 nsF	NMOS	TS	28	5	S9508A	AMI	40	MCM68369-25	Motorola	(4055)							
	350 nsF	NMOS	TS	28	5	R09508	GI		MCM68370-25	Motorola	(4055)							
						AM9240-35	AMD		IPD2364E-25	NEC	(4102)							
8000x8	350 nsF	NMOS	TS	28	5	TMS2564-35	TI (4229)		R2364B25	Rockwell								
	450 nsF	NMOS	TS	28	5	\$MJ2564-45 †	TI (4229)	45	2364-25	Signetics								
						TMS2564-45	TI (4229)		TMS2764H-25	TI	(4228)							
8192x4x2	450 nsF	NMOS	TS	24	5	2364B	SGS		TMM2364	Toshiba								
						26S64	SGS		TTL	TS	24	5						
						M2362	SGS		28	5	F3566-25	Fairchild (1466)						
8192x8	120 nsF	NMOS	TS	24	5	S68164	AMI	50	F3568-25	Fairchild (1466)								
	150 nsF	CMOS	TS	24	5	SCM23C64-3	SSS		F3569-25	Fairchild (1466)								
	200 nsF	CMOS	TS	24	5	SCM23C64-4	SSS		F3570-25	Fairchild (1467)								
		NMOS	TS	24	5	AM9265-20	AMD (3923)		2364-30	Signetics								
						R09464AD	GI		300 ns *	NMOS	TS	28	5					
						R09464D	GI		300 nsF	CMOS	TS	24	5					
						MCM68364	Motorola (4048,4054)	55	HCMP23C64C	Hughes								
						R2364A2	Rockwell		883/23C64M †	SSS								
						2664-20	Signetics (4198)		SCM23C64M †	SSS								
						2664A-20	Signetics		SCM23C65	SSS								
						TMM2366	Toshiba		23SC65	Supertex								
						28	5	23SC65A	Supertex		120							
						AM9264-20	AMD (3922)	60	G5364-3	GTEMicro								
						XLS2366	EXEL		G5365-3	GTEMicro								
						R09864AD	GI		SCM23C64	SSS								
						R09864D	GI		23SC64	Supertex								
									23SC64A	Supertex	125							
									NMOS	TS	24	5						
									AM9264C	AMD	(3922)							

† Military Temperature Range (-55° to 125°C) ns\* - Nanoseconds Typical nsF - Nanoseconds over Full Temperature Range nsR - Nanoseconds at Room Temperature OC - Open Collector

# IC MASTER

## MEMORY-ROMs (Cont'd)

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line
(Cont'd)									(Cont'd)								
8192x8	300 nsF	NMOS	TS	24	5	S4264	AMI	5	8192x8	450 nsF	CMOS	TS	24	5	MP2364-45	MicroPwr	70
						XLS2364	EXEL								MP2364C	MicroPwr	
						R09464AC	GI								MP2365-45	MicroPwr	
						R09464C	GI								G5364-4	GTEMicro	
						2364-3	GTEMicro	5							G5365-4	GTEMicro	
						MK36000-5	Mostek								AM9264B	AMD (3922)	
						MKB36000-84	† Mostek								XLS2364-450	EXEL	
						MKB37000-85	† Mostek								R09464AB	GI	
						MCM68364-30	Motorola								R09464B	GI	75
							(4048,4054)								MCM68367-45	Motorola (4054)	
						μPD2364-30	NEC (4102)	10							MCM68368-45	Motorola (4054)	
						R2364A3	Rockwell								MM52164	National	
						2664A-30	Signetics								2664A-45	Signetics	
						2664AE-30	† Signetics								2664AM-45	† Signetics	80
						SY2364-3	Syaortek (4211)								SY2364	Syaortek (4211)	
						SY2364A-3	Syaortek (4211)	15							SYM2364	† Syaortek (535)	
						SYM2364-3	† Syaortek (535)								SYM2364A	† Syaortek (535)	
						SYM2364A-3	† Syaortek (535)								TMS4764	TI (4225)	
						TMS4764-30	TI (4225)								TMS4764-45	TI (4225)	
						VT2364A	VTI (4250)								VT2364	VTI (4250)	85
					28	5	AM9265C	AMD (3923)	20						AM9265B	AMD (3923)	
						XLS2365	EXEL								XLS2365-450	EXEL	
						R09864AC	GI								XLS2366-450	EXEL	
						R09864C	GI								R09864AB	GI	90
						R2364B3	Rockwell								R09864B	GI	
						SY2365-3	Syaortek (4211)	25							μPD2364-45	NEC (4102)	
						SY2365A-3	Syaortek (4211)								2364-45	Signetics	
						SYM2365-3	† Syaortek (535)								SY2365	Syaortek (4211)	
						SYM2365A-3	† Syaortek (535)								SY2365A	Syaortek (4211)	95
						TMS2764H-30	TI (4228)								SYM2365	† Syaortek (535)	
						VT2365A	VTI (4251)	30							SYM2365A	† Syaortek (535)	
						VT2366A	VTI (4251)								TMS2764H-45	TI (4228)	
						HN61365	Hitachi								VT2365	VTI (4251)	100
						HN61366	Hitachi								VT2366	VTI (4251)	
						MP2364-35	MicroPwr								MP2366	MicroPwr	
						MP2365-35	MicroPwr	35		1000 nsF	CMOS	TS	24	1.5	MP2364LP1	MicroPwr	
						SMM2364	SMOS								MP2365LP1	MicroPwr	
					28	5	HN61364	Hitachi									
						SMM2365	SMOS										
						S68A364	AMI (3933)	40		1500 nsF	CMOS	TS	28	5	CM6400	Supertex	
						F3564-35	Fairchild (1465)										
						F3565-35	Fairchild (1465)										
						2364-4	GTEMicro										
						IM7364C	Intersil										
						MCM68365-35	Motorola										
							(4048,4054)										
						MCM68366-35	Motorola										
							(4048,4054)	45									
						MCM68A364	Motorola										
						MCM68A364P3	Motorola										
						μPD2364-35	NEC (4102)										
						TMS4764-35	TI (4225)										
						S2364A	AMI (3934)	50									
						MCM68369-35	Motorola (4055)										
						MCM68370-35	Motorola (4055)										
						2364-35	Signetics										
						2664AM-35	† Signetics										
						F3566-35	Fairchild (1466)	55									
						F3568-35	Fairchild (1466)										
						F3569-35	Fairchild (1466)										
						F3570-35	Fairchild (1467)										
						23MC65	Supertex										
						23MC65A	Supertex										
						23SC65-3	Supertex										
						23SC65A-3	Supertex										
					28	5	23MC64	† Supertex	60								
						23MC64A	† Supertex										
						23SC64-3	Supertex										
						23SC64A-3	Supertex	65									
										250 nsF	CMOS	TS	28	5	HN613128	Hitachi	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



**MEMORY-ROMs (Cont'd)**

Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line	Organization	Access Time (Max)	Type	Output	No. Pins	Supply Voltage, V	Device	Source	Line												
16384x8	250 nsF	CMOS	TS	28	5	(Cont'd)																							
						MP23128-25	MicroPwr	(1593)	5	250 nsF	CMOS	TS	28	5	MP2326-25	MicroPwr	(3936)												
	COM53128	RCA		SCM23256	SSS																								
							SCM23C128	SSS																					
		NMOS	TS	28	5	AM92128-25	AMD	(3924)	5	300 nsF	NMOS	—	28	5	AM92128-30	AMD	(3924)	15	300 nsF	NMOS	TS	28	5	AM92128-30	AMD	(3924)			
	AM92128D					AMD	(3924)	AM92128C							AMD	(3924)	RO9128C							GI		RO9128C	GI		
						S23128B	AMI	(3935)							23128-30	Signetics	(4202)												
						μPD23128	NEC	(4106)							SY23128-3	↑ Synartek	(4211)												
						23128-25	Signetics	(4202)							SY23128A-3	↑ Synartek	(4211)												
						TMS27128-25	TI	(4227)							SY23130-3	↑ Synartek	(4211)												
						TMS47128-25	TI	(4224)							SY23130A-3	↑ Synartek	(4211)												
						TMM27128-25	Toshiba								SYM23100A-3	↑ Synartek	(4211)												
						VT23128	VTI	(4252)							SYM23128-3	↑ Synartek	(535)												
						VT23129	VTI	(4252)							SYM23128A-3	↑ Synartek	(535)												
						RO9160	GI								SYM23128A-3	↑ Synartek	(535)												
						AM92128-30	AMD	(3924)							TMS27128-30	TI	(4227)												
						AM92128C	AMD	(3924)							S23128A	AMI	(3935)												
						300 nsF	TS	28	5						S3630A	AMI													
						350 nsF	TS	28	5						TMS47128-35	TI	(4224)												
						450 nsF	TS	28	5						AM92128-45	↑ AMD	(3924)												
															AM92128B	AMD	(3924)												
														RO9128B	GI														
														32128-45	Signetics														
														SY23128A	↑ Synartek	(4211)													
														SY23130	↑ Synartek	(4211)													
														SY23130A	↑ Synartek	(4211)													
														SYM23128	↑ Synartek	(535)													
														SYM23128A	↑ Synartek	(535)													
														TMS27128-45	TI	(4227)													
														TMS47128-45	TI	(4224)													
														VT23128B	VTI	(4252)													
														VT23129B	VTI	(4252)													
														CM1310	Supertex														
														μPD73128	NEC	(4114)													
														HN43128	Hitachi														
														HN43128	Hitachi														
														TMM23256	Toshiba														
														S23256C	AMI	(3936)													
														MCM63256-15	Motorola	(4048,4055,4056)													
														AM92256-20	AMD	(3925)													
														XLS23256-200	EXEL														
														XLS23257-200	EXEL														
														XLS23258-200	EXEL														
														MK38000-20	Mostek														
														MCM63256-20	Motorola	(4048,4055,4056)													
														23256A-20	Signetics														
														SY23256-2	↑ Synartek	(4211)													
														SY23256A-2	↑ Synartek	(4211)													
														COM53256	RCA	(1593)													
														XLS23C256	EXEL														
														MB83256	Fujitsu														
														HN613256	Hitachi														
														MP2325-25	MicroPwr														
														(Continued)															

† Military Temperature Range (-55° to 125°C)   ns\*—Nanoseconds Typical   nsF—Nanoseconds over Full Temperature Range   nsR—Nanoseconds at Room Temperature  
 OC—Open Collector   TS—Three-State   OE—Open Emitter

# IC MASTER

## MEMORY—Shift Registers

Bits Per Register	No. of Reg.	Operation	Process	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line	Bits Per Register	No. of Reg.	Operation	Process	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line
<b>Dynamic</b>									4	1	PP	Bipolar	20 M	5			(Cont'd)
100	2	SS	PMOS	2.0 M	±5	AM1406 AM1407 AM1506 AM1507	† AMD								RM8270 RM8271	† Raytheon	
256	4	SS	PMOS	5.0 M	±5	AM1402AC AM1402AM	AMD	5					25 M	5	54178 54179 54194 5495 74178 74179 74194 7495	† Fairchild	65
				10.0 M	±5	AM2802C AM2802M	AMD								74178 74179 74194 7495	Fairchild	
455	2	SS	CCD	20.0 M	17	CCD321	Fairchild										
512	1	SS	PMOS	2.0 M	±5	AM1405A AM2805M AM2807M	AMD	10							ZN5495A ZN7495A HD74194 HD7495A	† Ferranti	
				3.0 M	±5		† AMD										
				4.0 M	±5	AM2805C AM2807C	AMD								HD74194 HD74LS194A	Hitachi	70
	2	SS	PMOS	5.0 M	±5	AM1403AC AM1403AM	AMD	15							MC74194 DM54194	Motorola	
				10.0 M	±5	AM2803C AM2803M	AMD								DM5495 DM74194	† National	
1024	1	SS	PMOS	3.0 M	±5	AM2806M AM2808M	† AMD	20							DM7495 54194	National	80
				4.0 M	±5	AM2806C AM2808C	AMD								5495A 74194	† Signetics	(861)
				5.0 M	±5	AM1404AC AM1404AM	AMD	25							7495A 74LS95B	Signetics	(860)
				10.0 M	±5	AM2804C AM2804M	AMD								SN54178 SN54179	† TI	(861)
	2	SS	NMOS	1.0 M	5	AM2401 AM9401C AM9401M	AMD	30							SN54194 SN5495A	TI	(984)
				2.0 M	5		AMD								SN54L95	TI	(985)
				PMOS	2.0 M	MM4025 MM4026	† AMD								SN54199	† TI	(990)
				5.0 M	-10.5,5	AM2825M AM2826M	† AMD	35							SN54LS194A	† TI	(954)
				6.0 M	-10.5,5	AM2825C AM2826C	AMD								SN54LS95B	† TI	(954)
2048	1	SS	PMOS	2.0 M	-12.5	MM4027	† AMD	40					30 M	5	54195	† Fairchild	100
				4.0 M	-10.5,5	AM2827M	† AMD								54LS194A	Fairchild	
				6.0 M	-10.5,5	AM2827C	AMD								54LS195A	† Fairchild	
4096	4	PP	CCD	5.0 M	±5,12	CCD460	Fairchild								54LS295A	† Fairchild	
65536	1	SS	CCD	25 M	±5,12	F464	Fairchild	40							54LS395	† Fairchild	
<b>Static</b>																	
1-134	4	SS	NMOS	1 M	5	SR5015	SMC								54LS95B	† Fairchild	105
4	1	PP	Bipolar	3 M	5	SN74L95 SN74L99	TI								74195	Fairchild	
					12	375A/C	TeledyneS								74LS194A	Fairchild	
					15	375B/M	TeledyneS	45							74LS195A	Fairchild	
				6 M	5	DM54L95 DM74L95	† National								74LS295A	Fairchild	110
				15 M	5	93L00C 93L00M MC8300 MC9300	Fairchild	50							74LS395	Fairchild	
						M8271	Signetics	(876)							74LS95B	Fairchild	
						S8270	† Signetics								9300C	Fairchild	
						S8271	† Signetics								9300M	† Fairchild	
				20 M	5	MC7270 MC7271 MC8270 MC8271 RC8270 RC8271	Motorola	55							HD74195	Hitachi	115
							Raytheon								HD74LS195A	Hitachi	
							Raytheon	60							MC74195	Motorola	
															SN54LS194A	† Motorola	
															SN54LS195A	† Motorola	
															SN54LS295A	† Motorola	
															SN54LS395	† Motorola	120
															SN54LS95B	† Motorola	
															SN74LS194A	Motorola	
															SN74LS195A	Motorola	
															SN74LS295A	Motorola	125
															SN74LS395	Motorola	
															SN74LS95B	Motorola	
															DM54195	† National	
															DM74195	National	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

Master Selection Guide

MEMORY

MEMORY—Shift Registers (Cont'd)

Bits Per Register	No. of Reg.	Operation	Process	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line	
<b>Static</b>									
4	1	PP	Bipolar	30 M	5		(Cont'd)		
						DM8300	National	(Cont'd)	
						DM9300	† National		
						T54LS395	† SGS		
						T74LS395	SGS		
						54LS195A	† Signetics (530)	5	
						74195	Signetics (861)		
						74LS195A	Signetics		
						SN54195	† TI (991)		
						SN54LS195A	† TI (991)		
						SN54LS295B	† TI (1014)	10	
						SN54LS395A	† TI (1031)		
						SN74195	TI (991)		
						SN74LS195A	TI (991)		
						SN74LS295B	TI (1014)		
						SN74LS395A	TI (1031)	15	
				31 M	5	MC7495A	Motorola		
				35 M	5	MC4012	Motorola		
						MC4312	† Motorola		
				70 M	5	54S194	† Fairchild	20	
						74S194	Fairchild		
						93S00C	Fairchild		
						93S00M	† Fairchild		
						74S194	Signetics	85	
						74S195	Signetics		
						SN54S194	† TI (990)	25	
						SN54S195	† TI (991)		
						SN74S194	TI (990)		
						SN74S195	TI (991)		
				140 M	-5.2	95H00C	Fairchild		
				150 M	-5.2	F1000C	Fairchild	30	
						MC10141	Motorola		
						MC10541	† Motorola		
						10141	Signetics (858)		
				250 M	-5.2	F10141	Fairchild		
			CMOS	2 M	3-15	F40194BM	† Fairchild	35	
						F40195BM	† Fairchild		
						F4035BM	† Fairchild		
						HD14035B	Hitachi		
						MC14035BC	Motorola		
						CD4035BC	National	40	
						MM54C195	† National		
						CD4035BE	RCA (838)		
						883/4035B	† SSS		
						SCL4035B	SSS		
						CM4035AE	Solitron	45	
				3 M	3-15	F40194BC	Fairchild		
						F40195BC	Fairchild		
						F4035BC	Fairchild		
						MC14035BA	† Motorola	50	
						CD4035BM	† National		
						MM74C195	National		
						CD4035B	† RCA (838)		
						CM4035A	† Solitron		
				4.5 M	3-15	MC14194BA	† Motorola		
				6.5 M	3-15	MM54C95	† National	55	
						MM74C95	National		
			PS	Bipolar	10 M	5	5494	† Fairchild	60
						7494	Fairchild		
						ZN5494	† Ferranti		
						ZN7494	Ferranti		
						MC7494	Motorola		
						5494	Signetics (860)		
						7494	Signetics (860)		
							(Continued)		
4	1	PS	Bipolar	10 M	5		(Cont'd)		
						SN5494	† TI (953)	65	
						SN7494	TI (953)		
		SP		325 M	-5.2	MC1694	† Motorola		
		SS		0.5 M	15	MC686	Motorola		
	2	SP	CMOS	1 M	5	CM4015AE	Solitron	70	
				1.5 M	5	CM4015A	† Solitron		
				2.5 M	3-15	F4015BC	Fairchild		
						HD14015B	Hitachi		
						MC14015BC	Motorola		
						CD4015A	National		
						CD4015BE	RCA (838)		
				3 M	3-15	F4015BM	† Fairchild	75	
						CD4015B	† RCA (838)		
						883/4015B	† SSS		
						SCL4015B	SSS		
						MC14015BA	† Motorola		
5	1	PP	Bipolar	10 M	5	5496	† Fairchild	80	
						7496	Fairchild		
						ZN5496	† Ferranti		
						ZN7496	Ferranti		
						HD7496	Hitachi		
						MC7496	Motorola	85	
						DM5496	† National		
						DM7496	National		
						5496	† Signetics (528,860)		
						54LS96	Signetics (529)	90	
						7496	Signetics (860)		
						74LS96	Signetics		
						SN5496	† TI (954)	95	
						SN7496	TI (954)		
				25	5	SN54LS96	† TI (954)		
						SN74LS96	TI (954)		
	4	SS	CMOS	1.5 M	5	CM4006AD	† Solitron	100	
				2 M	3-15	F4006BC	Fairchild		
						CD4006BE	RCA (838)		
						CM4006A	† Solitron		
						CM4006AE	Solitron		
				2.5 M	3-15	F4006M	† Fairchild	105	
						CD4006A	National		
				3 M	3-15	883/4006B	† SSS		
						SCL4006B	SSS		
				4 M	5-15	MC14006BC	Motorola	110	
				7 M	5-15	MC14006BA	† Motorola		
8	1	PP	Bipolar	15 M	5	DM7546	† National	115	
						DM8546	National		
				25 M	5	54198	† Fairchild	120	
						54199	† Fairchild		
						74198	Fairchild		
						74199	Fairchild		
						HD74198	Hitachi		
						HD74199	Hitachi		
						DM54198	† National		
						DM54199	† National		
						DM74198	National		
						DM74199	National		
						74199	Signetics (861)		
						SN54198	† TI (992)		
						SN54199	† TI (993)		
						SN74198	TI (992)		
						SN74199	† TI (993)		
				35 M	5	AM25LS23C	AMD (1427)	(Continued)	

† Military Temperature Range (-55° to 125°C)  
\* Typical Values

SS = Serial In, Serial Out

PP = Parallel In, Parallel Out

IC MASTER

MEMORY-Shift Registers (Cont'd)

Bits Per Register	No. of Reg.	Operation	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line	Bits Per Register	No. of Reg.	Operation	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																										
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<td></td> <td></td> <td></td> <td></td> <td></td> <td>SN54LS299 †TI (1016)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN54LS323 †TI (1018)</td> <td>10</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN64LS299 TI</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN74LS323 TI (1018)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td>50 M</td> <td>5</td> <td>SN54S299 †TI (1016)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN74S299 TI (1016)</td> <td>15</td> </tr> <tr> <td></td> <td></td> <td></td> <td>CMOS</td> <td>2.5 M</td> <td>3-15</td> <td>SCL4034B SSS</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td>3-18</td> <td>MC14034BC Motorola</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td>5</td> <td>CM4034AE Solitron</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td>3 M</td> <td>3-15</td> <td>CD4034BC National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4034BM National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4034B †RCA (838,840)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>883/4034B †SSS</td> <td>20</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>TC4034B Toshiba</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td>3-18</td> <td>HD14034B Hitachi</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>MC14034BA Motorola</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td>5</td> <td>CM4034A †Solitron</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td>PS</td> <td>Bipolar</td> <td>14 M</td> <td>5</td> <td>DM7590 †National</td> <td>25</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>DM8590 National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td>20 M</td> <td>5</td> <td>54165 †Fairchild</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>74165 Fairchild</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>MC74165 Motorola</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>DM54165 †National</td> <td>30</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>DM74165 National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>54165 †Signetics (861)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>74165 Signetics (861)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN54165 †TI (979)</td> <td></td> </tr> <tr> <td></td> <td></td> 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<td></td> <td></td> <td>DM74166 National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>54166 †Signetics (861)</td> <td>45</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>74166 Signetics (861)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN54166 †TI (979)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN54LS165 †TI (979)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN54LS166 †TI (979)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN74166 TI (979)</td> <td>50</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN74LS165 TI (979)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SN74LS166 TI (979)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td>CMOS</td> <td>1 M</td> <td>3-15</td> <td>F4014BC Fairchild</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>F4021BC Fairchild</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td>2.5 M</td> <td>3-15</td> <td>HD14021B Hitachi</td> <td>55</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>MC14021BC Motorola</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4014A National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4021A National</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4014BE RCA (838)</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4021BE RCA (838)</td> <td>60</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SCL4014B SSS</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>SCL4021B SSS</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CM4018AE Solitron</td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CM4021AE Solitron</td> <td></td> </tr> </tbody> </table>								Static		(Cont'd)		8	1	PP	Bipolar	35 M	5	(Cont'd)								AM25LS323M †AMD (1427)								SN54LS323 †AMD								SN74LS323 AMD								54LS299 †Fairchild								54LS323 †Fairchild	5							74LS299 Fairchild								74LS323 Fairchild								SN54LS299 †TI (1016)								SN54LS323 †TI (1018)	10							SN64LS299 TI								SN74LS323 TI (1018)						50 M	5	SN54S299 †TI (1016)								SN74S299 TI (1016)	15				CMOS	2.5 M	3-15	SCL4034B SSS							3-18	MC14034BC Motorola							5	CM4034AE Solitron						3 M	3-15	CD4034BC National								CD4034BM National								CD4034B †RCA (838,840)								883/4034B †SSS	20							TC4034B Toshiba							3-18	HD14034B Hitachi								MC14034BA Motorola							5	CM4034A †Solitron					PS	Bipolar	14 M	5	DM7590 †National	25							DM8590 National						20 M	5	54165 †Fairchild								74165 Fairchild								MC74165 Motorola								DM54165 †National	30							DM74165 National								54165 †Signetics (861)								74165 Signetics (861)								SN54165 †TI (979)								SN74165 TI (979)	35					25 M	5	54166 †Fairchild								54LS165 †Fairchild								74166 Fairchild								74LS165 Fairchild								HD74166 Hitachi	40							SN54LS165 †Motorola								SN74LS165 Motorola								DM54166 †National								DM74166 National								54166 †Signetics (861)	45							74166 Signetics (861)								SN54166 †TI (979)								SN54LS165 †TI (979)								SN54LS166 †TI (979)								SN74166 TI (979)	50							SN74LS165 TI (979)								SN74LS166 TI (979)					CMOS	1 M	3-15	F4014BC Fairchild								F4021BC Fairchild						2.5 M	3-15	HD14021B Hitachi	55							MC14021BC Motorola								CD4014A National								CD4021A National								CD4014BE RCA (838)								CD4021BE RCA (838)	60							SCL4014B SSS								SCL4021B SSS								CM4018AE Solitron								CM4021AE Solitron		<table border="1"> <thead> <tr> <th colspan="2"></th> <th colspan="2">(Cont'd)</th> </tr> </thead> <tbody> <tr> <td>8</td> <td>1</td> <td>PS</td> <td>CMOS</td> <td>3 M</td> <td>3-15</td> <td>F4014BM †Fairchild</td> <td>(Cont'd)</td> <td>65</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>F4021BM †Fairchild</td> <td></td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>MC14014BA †Motorola</td> <td></td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>MC14021BA †Motorola</td> <td></td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4014B †RCA (838)</td> <td></td> <td>70</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>CD4021B †RCA (838)</td> <td></td> <td></td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>883/4014B †SSS</td> <td></td> <td></td> </tr> 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† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

MEMORY—Shift Registers (Cont'd)

Bits Per Register	No. of Reg.	Operation	Process	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line	Bits Per Register	No. of Reg.	Operation	Process	Frequency (Hz - Spec)	Supply Voltage, V	Device	Source	Line		
<b>Static (Cont'd)</b>									<b>(Cont'd)</b>										
8	1	SS	CMOS	2.5 M	3-15	CD4094BE 883/4094B SCL4094B TC4094B	RCA † SSS SSS Toshiba	(838)	64	4	SS	PMOS	1.5 M	-12.5	3342	Fairchild	(Cont'd)		
	2	SS	Bipolar	10 M	5	93L28C 93L28M MC8328 MC9328	Fairchild † Fairchild Motorola † Motorola		80	4	SS	NMOS	1 M 3 M	5 5	SR5015-80 M142 M142A	SMC SGS SGS			
				15 M	5	RC8277 RM8277 N8277	Raytheon † Raytheon Signetics					PMOS	1.5 M	-12.5	AM3347 3347	AMD Fairchild	65		
				20 M	5	9328C 9328M	Fairchild † Fairchild		5				2 M	-12.5	3357-2	Fairchild			
10	1	PP	Bipolar	25 M	5	RC8274 RM8274 S8274	Raytheon † Raytheon † Signetics		10				2.5 M	-12.5	AM2847M F2847LM F2847M	† AMD † Fairchild † Fairchild	70		
		SP	Bipolar	25 M	5	RC8273 RM8273 N8273 S8273	Raytheon † Raytheon Signetics † Signetics	(876)	15	81	4	SS	NMOS	1 M	5	SR5015-81 SR5018	SMC SMC		
16	1	S/SP	Bipolar	20M	5	54F673 54F674 54F675 54F676 74F673 74F674 74F675 74F676 74F677 74F678 74F679 74F680	Fairchild Fairchild Fairchild Fairchild Fairchild Fairchild Fairchild Fairchild Signetics Signetics Signetics Signetics	(763)	20	15	96	4	SS	PMOS	2.5 M 3 M	-12.5 -12.5	AM2896M AM2896C	† AMD AMD	
		SP	NMOS	3.0M	5	TSC9403 TSC9404	TeledyneS TeledyneS	(3101) (3101)	25	128	1	SS	CMOS	5 M	12	MS625	RTC		
16-128	1	SS	CMOS	1.5 M 3 M	3-15 3-15	MC14562BC MC14562BA	Motorola † Motorola		30	2	SS	PMOS	1 M 2 M	-12.5 -12.5	MK1002 AM2809M AM2810C AM2810M TMS3114	AMD AMD AMD AMD AMD	80		
30	1	SS	CMOS	1 M	3-15	MD4330B	Mitel		35				2.5 M	-12.5	AM2809C AM2814C AM2814M	AMD AMD † AMD	85		
32	6	SS	PMOS	1 M	-12.5	3348 3349	Fairchild Fairchild		40	4	SS	PMOS	2.5 M	-12.5	AM2855 AM4055 AM5055	AMD † AMD AMD	90		
64	1	SS	CMOS	1 M	3-15	F4031BC CD4031BE F4031BM CD4031B F4557BM MC1455BC CD4031BC CD4031BM F4557BC MC14557BA MS612	Fairchild RCA † Fairchild † RCA † Fairchild Motorola National † National Fairchild † Motorola RTC	(838) (838)	50	133	4	SS	NMOS	1 M	5	SR5015-133 SR5017	SMC SMC		
	2	SS	CMOS	4 M	3-15	MC14517BC CD4517BE SCL4517B MC14517BA CD4517BE 883/4517B	Motorola † RCA SSS † Motorola † RCA † SSS	(838)	30	256	1	SS	TTL	30 M	5	TDC1006 TDC1006	† TRW-LSI TRW-LSI	95	
				5 M	3-15 3-18	F4557BC MC14557BA	Fairchild † Motorola		45	2	SS	PMOS	2.5 M	-12.5	AM2856 AM4056 AM5056	AMD † AMD AMD			
				10 M	3-16	MS612	RTC		55	512	1	SS	PMOS	2.5 M	-12.5	AM2857C AM4057 AM5057	AMD † AMD AMD	100	
	4	SS	CMOS	4 M	4.5-12.5	F4731BC F4731BM	Fairchild † Fairchild		60	1024	1	SS	PMOS	1.5 M 2 M	-12.5 -12.5	AM2533 AM2833C AM2833M	AMD AMD † AMD		
			TTL	30 M	5	TDC1005 TDC1005	† TRW-LSI TRW-LSI		92304	1	SS	Bubble	0.1 M	—	TIB0203S	TI	105		
									290560	1	SS	Bubble	—	—	H4701B	Hitachi			

† Military Temperature Range (-55° to 125°C)  
\* Typical Values

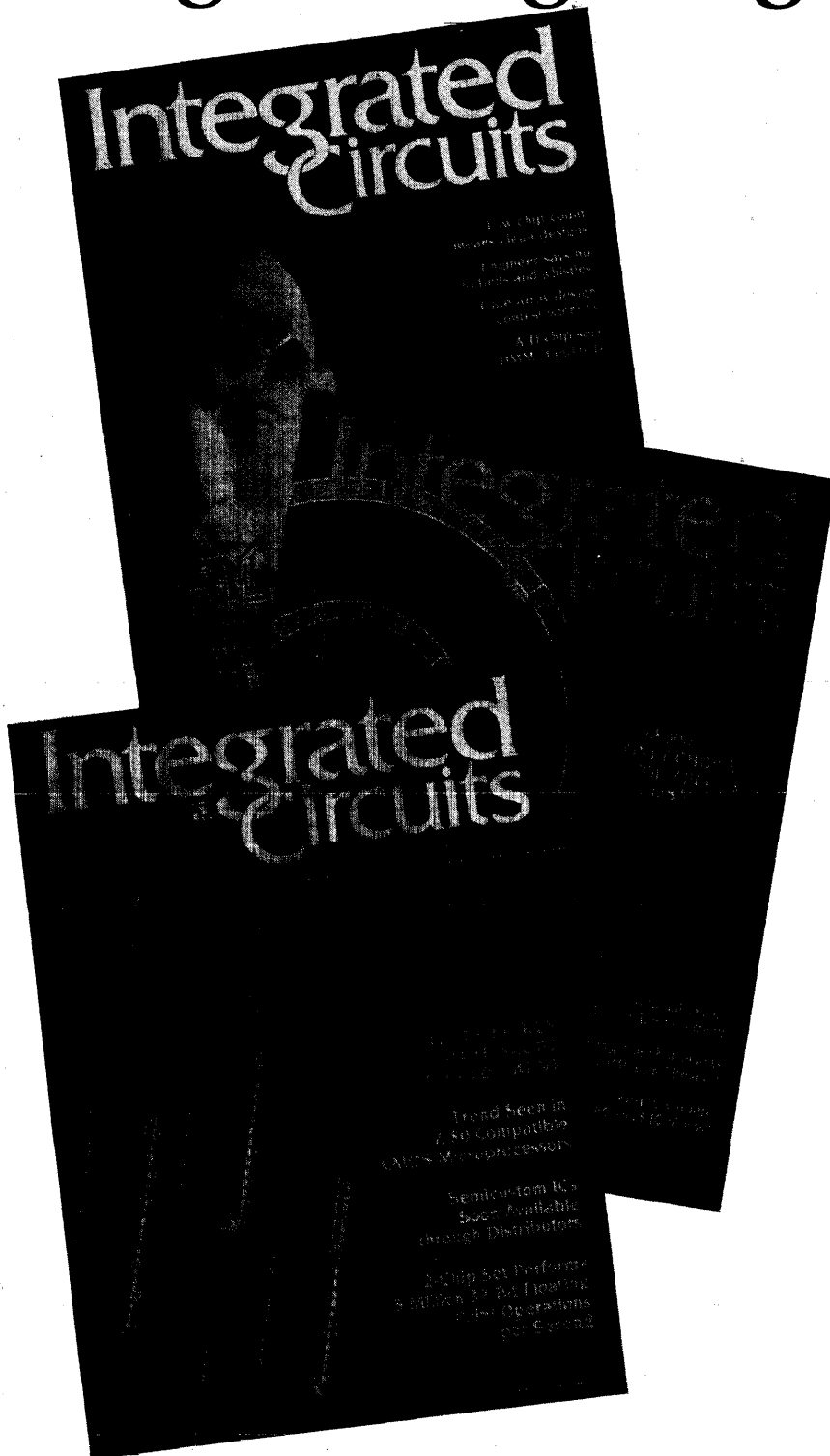
SS = Serial In, Serial Out

PP = Parallel In, Parallel Out

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# Integrated Circuits Magazine

## The magazine for the engineering design strategist\*



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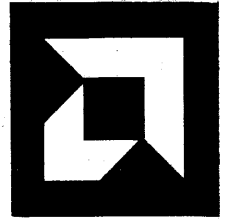
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# Advanced Micro Devices



## Bipolar Memories Am27S35A/35/37A/37

### DISTINCTIVE CHARACTERISTICS

- Member of AMD's Generic Family of 8-bit wide registered PROMs
- On-chip edge-triggered registers – ideal for pipelined microprogrammed systems
- Versatile synchronous and asynchronous enables for simplified word expansion
- Versatile programmable register INITIALIZE either asynchronous (Am27S35A/35) or synchronous (Am27S37A/37)
- Slim, 24-pin, 300-mil lateral center package occupies approximately 1/3 the board space required by standard discrete PROM and register
- Consumes approximately 1/2 the power of separate PROM/register combination for improved system reliability
- Fast standard version – 40ns max setup and 25ns max clock-to-output allows system speed improvements
- "A" version offers improved AC performance in critical paths (35ns max setup and 20ns max clock-to-output)
- Platinum-Silicide fuses guarantee high reliability, fast programming, and exceptionally high programming yields (typ > 98%)
- AC performance is factory tested utilizing programmed test words and columns

### GENERIC SERIES CHARACTERISTICS

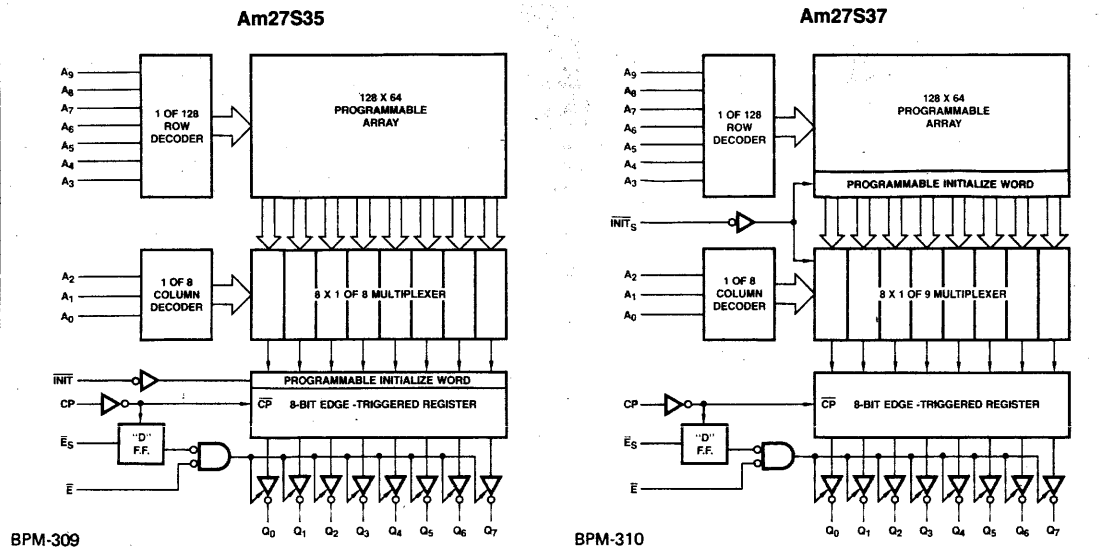
The Am27S35A/35 and Am27S37A/37 are members of an Advanced PROM series incorporating common electrical characteristics and programming procedures. All parts in this series are produced with a fusible link at each memory location storing a logic LOW and can be selectively programmed to a logic HIGH by applying appropriate voltages to the circuit.

All parts are fabricated with AMD's fast programming highly reliable Platinum-Silicide Fuse technology. Utilizing easily implemented programming (and common programming equipment) these products can be rapidly programmed to any customized pattern. Extra test words are preprogrammed during manufacturing to insure extremely high field programming yields and produce excellent parametric correlation.

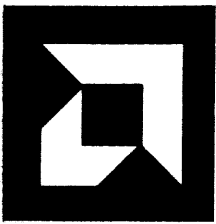
Platinum-Silicide was selected as the fuse link material to achieve a well controlled melt rate resulting in large nonconductive gaps that ensure long-term reliability. Extensive operating testing has shown that this low-field, large-gap technology offers the best reliability for fusible link PROMs.

Common design features include active loading of all critical AC paths regulated by a built-in temperature and voltage-compensated bias network to provide excellent parametric performance over the full military power supply and temperature ranges. Selective feedback techniques have been employed to minimize delays through all critical paths producing the fastest speeds possible from Schottky processed PROMs.

### BLOCK DIAGRAMS



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# Advanced Micro Devices

## Bipolar Memories Am27S65/75/85

### SERIAL SHADOW REGISTER (SSR) DIAGNOSTICS CAPABILITIES

- Serial access to output register to allow input of diagnostic control information
- Serial access of output register allows observation of register data
- Eliminates the need for diagnostics code internal to the PROM, allowing increased applications code density
- Simplified diagnostics increases system reliability
- Separate diagnostic register allows real time "snap shot" of machine state

### DISTINCTIVE CHARACTERISTICS

- Highest density fastest performance PROM organization
- On-chip edge-triggered registers – ideal for pipelined microprogrammed systems
- On-chip diagnostic shift register for serial observability and controllability of the output register
- Fast standard version: 30ns max setup and 15ns max clock-to-output allows system speed improvements
- "A" version offers improved AC performance in critical paths (25ns max setup and 12ns max clock-to-output)
- User-programmable synchronous and asynchronous Enables
- User-programmable for synchronous or asynchronous Initialize
- Increased drive capability, 24mA I<sub>OL</sub>
- THINDIP, 24-pin, 300-mil lateral center package increases overall board density
- Platinum-Silicide fuses guarantee high reliability, fast programming and exceptionally high programming yields (typ >98%)
- AC performance is factory tested utilizing programmed test rows and columns

### GENERIC SERIES CHARACTERISTICS

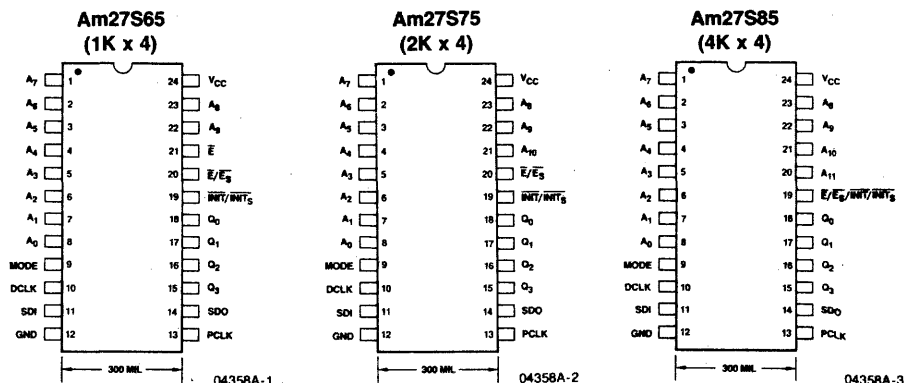
The Am27S65, Am27S75, and Am27S85 are members of an advanced PROM series incorporating common electrical characteristics and programming procedures. All parts in this series are produced with a fusible link at each memory location storing a logic LOW which can be selectively programmed to a logic HIGH by applying appropriate programming voltages to the circuit.

All parts are fabricated with AMD's fast-programming highly reliable Platinum-Silicide Fuse technology. Utilizing easily implemented programming algorithms (and common programming equipment), these products can be rapidly programmed to any customized pattern. Extra test words are preprogrammed during manufacturing to insure extremely high field programming yields and produce excellent AC and DC parametric correlation.

Platinum-Silicide was selected as the fuse link material to achieve a well controlled melt rate resulting in large nonconductive gaps that ensure very stable long-term reliability. Extensive operating testing has proven that this low-field, large-gap technology offers the best reliability for fusible link PROMs.

Common design features include active loading of all critical AC paths, which are regulated by a built-in temperature- and voltage-compensated bias network to provide excellent parametric performance over the full military power supply and temperature ranges. Selective feedback techniques have been employed to minimize delays through all critical paths producing the fastest speeds possible from Schottky processed PROMs.

### 4-WIDE REGISTERED PROMs WITH SSR DIAGNOSTICS CONNECTION DIAGRAMS



Note: Pin 1 is marked for orientation.

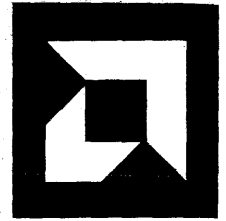
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Order # 04358A

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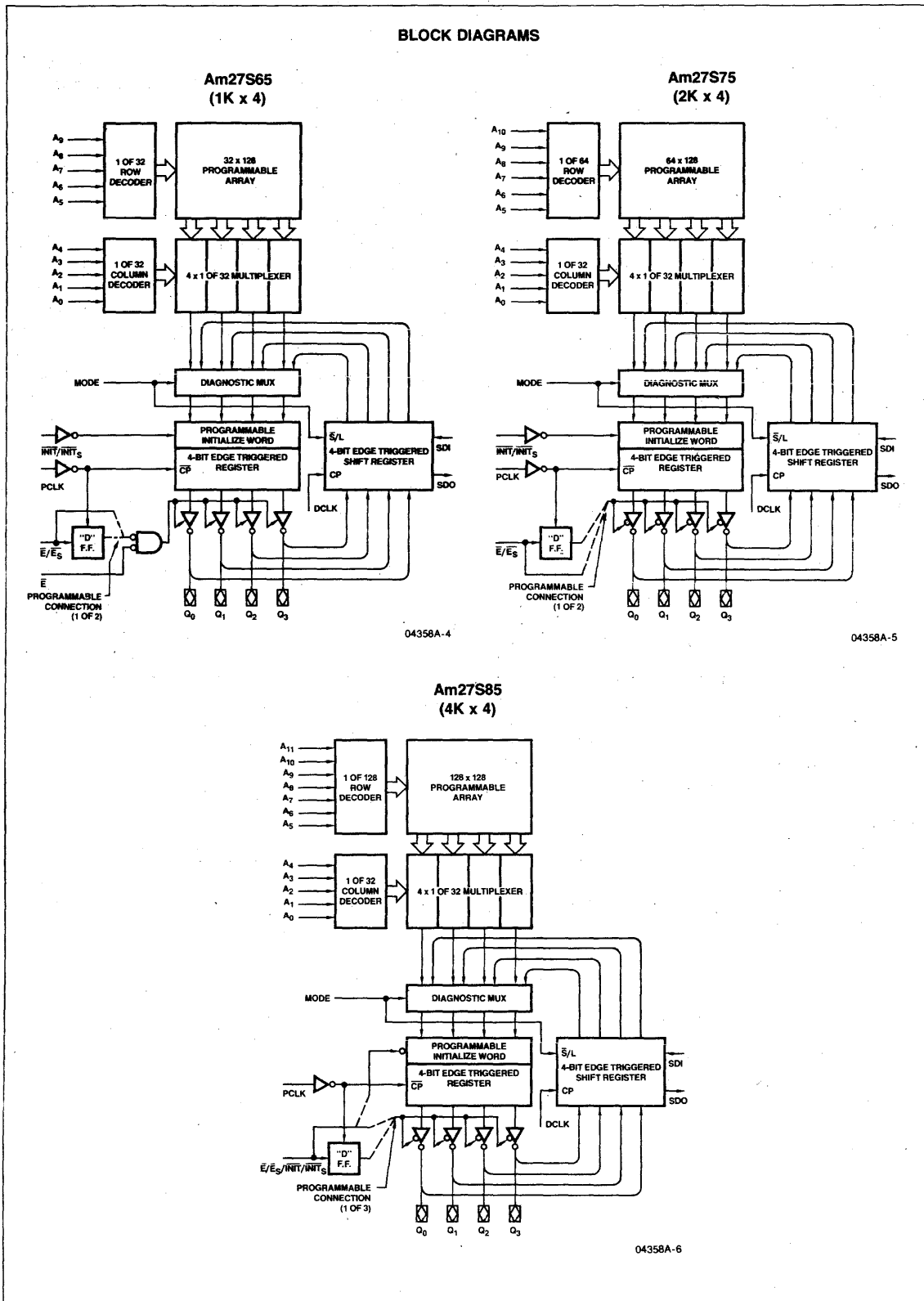


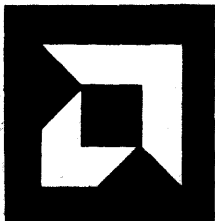
# Advanced Micro Devices



## Bipolar Memories Am27S65/75/85

### BLOCK DIAGRAMS





# Advanced Micro Devices

## Bipolar Memories Am27S45A/45/47A/47

### DISTINCTIVE CHARACTERISTICS

- Member of AMD's Generic Family of 8-bit wide registered PROMs
- On-chip edge-triggered registers – ideal for pipelined microprogrammed systems
- User programmable for synchronous or asynchronous enable for simplified word expansion
- Versatile programmable register INITIALIZE either asynchronous (Am27S45A/45) or synchronous (Am27S47A/47)
- Slim, 24-pin, 300-mil lateral center package occupies approximately 1/3 the board space required by standard discrete PROM and register
- Consumes approximately 1/2 the power of separate PROM/register combination for improved system reliability
- Fast standard version – 45ns max setup and 25ns max clock-to-output allows system speed improvements
- "A" version offers improved AC performance in critical paths (40ns max setup and 20ns max clock-to-output)
- Platinum-Silicide fuses guarantee high reliability, fast programming, and exceptionally high programming yields (typ > 98%)
- AC performance is factory tested utilizing programmed test words and columns

### GENERIC SERIES CHARACTERISTICS

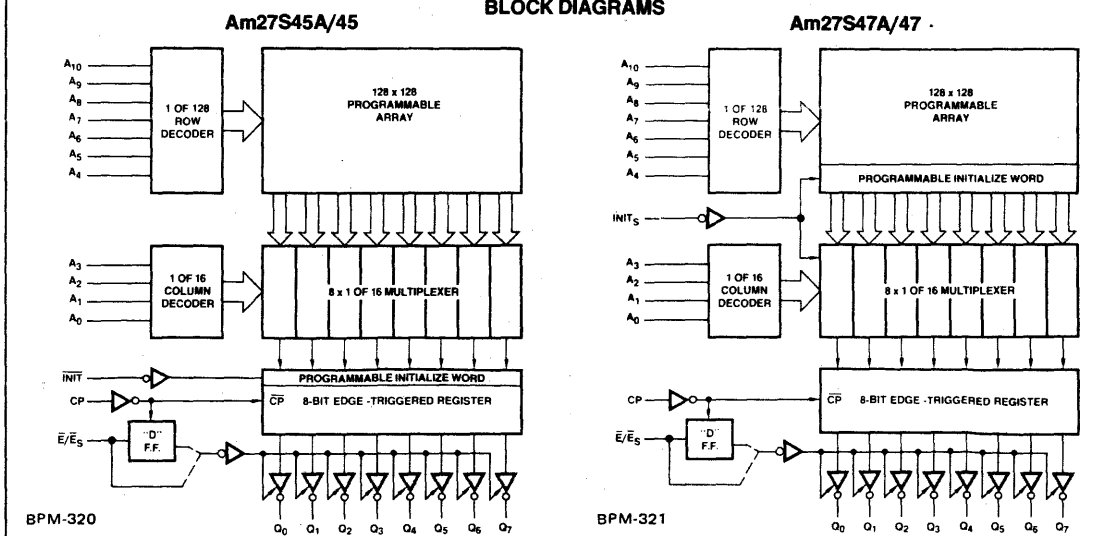
The Am27S45A/45 and Am27S47A/47 are members of an Advanced PROM series incorporating common electrical characteristics and programming procedures. All parts in this series are produced with a fusible link at each memory location storing a logic LOW which can be selectively programmed to a logic HIGH by applying appropriate programming voltages to the circuit.

All parts are fabricated with AMD's fast programming highly reliable Platinum-Silicide Fuse technology. Utilizing easily implemented programming algorithms (and common programming equipment) these products can be rapidly programmed to any customized pattern. Extra test words are preprogrammed during manufacturing to insure extremely high field programming yields and produce excellent AC and DC parametric correlation.

Platinum-Silicide was selected as the fuse link material to achieve a well controlled melt rate resulting in large nonconductive gaps that ensure very stable long-term reliability. Extensive operating testing has proven that this low-field, large-gap technology offers the best reliability for fusible link PROMs.

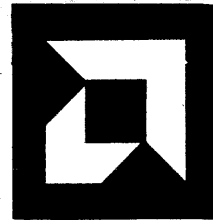
Common design features include active loading of all critical AC paths, which are regulated by a built-in temperature and voltage compensated bias network to provide excellent parametric performance over the full military power supply and temperature ranges. Selective feedback techniques have been employed to minimize delays through all critical paths producing the fastest speeds possible from Schottky processed PROMs.

### BLOCK DIAGRAMS



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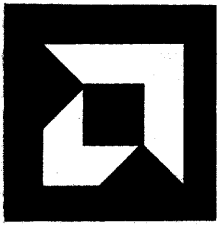


## Bipolar Memories Am27S180A/181A/280A/281A/180/181/280/281

### DISTINCTIVE CHARACTERISTICS

Part Number	Package Width	Other Features
Am27S180A	24-Pin, Plug in Replacement for Industry Standard 600-mil Configuration No Board Changes Required	Ultra fast – 35ns max
Am27S181A		Fast – 60ns max
Am27S180		
Am27S181		
Am27S280A	New Space-Saving 24-Pin, THINDIP, 300-mil Configuration Increases Overall Board Density	Ultra fast – 35ns max
Am27S281A		Fast – 60ns max
Am27S280		
Am27S281		

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## Bipolar Memories Am27S180A/181A/280A/281A/180/181/280/281

### DISTINCTIVE CHARACTERISTICS

- Platinum-Silicide fuses guarantee high reliability, fast programming and exceptionally high programming yields (typ > 98%)
- AC performance is factory tested utilizing programmed test words and columns
- Voltage and temperature compensated providing extremely flat AC performance over military range
- Members of Generic PROM series utilizing standard programming algorithm

### GENERIC SERIES CHARACTERISTICS

These 8K PROMs are members of an Advanced PROM series incorporating common electrical characteristics and programming procedures. All parts in this series are produced with a fusible link at each memory location storing a logic LOW and can be selectively programmed to a logic HIGH by applying appropriate voltages to the circuit.

All parts are fabricated with AMD's fast programming highly reliable Platinum-Silicide Fuse technology. Utilizing easily implemented programming (and common programming equipment) these products can be rapidly programmed to any customized pattern. Extra test words are pre-programmed during manufacturing to insure extremely high field programming yields and produce excellent parametric correlation.

Platinum-Silicide was selected as the fuse link material to achieve a well controlled melt rate resulting in large non-conductive gaps that ensure very stable long-term reliability. Extensive operating testing has proven that this low-field, large-gap technology offers the best reliability for fusible link PROMs.

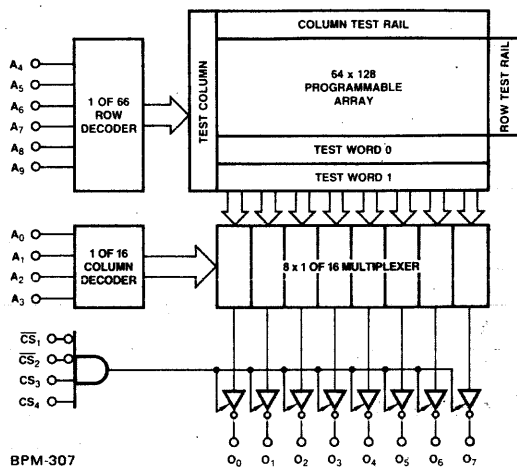
Common design features include active loading of all critical AC paths regulated by a built-in temperature and voltage compensated bias network to provide excellent parametric performance over MIL supply and temperature ranges. Selective feedback techniques have been employed to minimize delays through all critical paths.

These PROMs are manufactured using Advanced Micro Devices' selective oxidation process, IMOX. This advanced process combined with a merged fuse array permits an increase in density and a decrease in internal capacitance resulting in the fastest possible PROMs.

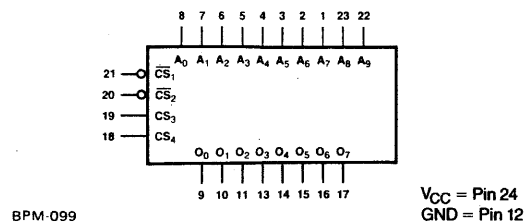
### FUNCTIONAL DESCRIPTION

These 8K PROMs are high speed electrically programmable Schottky read only memories. Organized in the industry standard 1024 x 8 configuration, they are available in both open collector (Am27S180A/180 and Am27S280A/280) and three-state (Am27S181A/181 and Am27S281A/281) output versions. After programming, stored information is read on outputs O<sub>0</sub>-O<sub>7</sub> by applying unique binary addresses to A<sub>0</sub>-A<sub>9</sub> and holding  $\overline{CS}_1$  and  $\overline{CS}_2$  LOW and  $\overline{CS}_3$  and  $\overline{CS}_4$  HIGH. All other valid input conditions on  $\overline{CS}_1$ ,  $\overline{CS}_2$ ,  $\overline{CS}_3$  and  $\overline{CS}_4$  place O<sub>0</sub>-O<sub>7</sub> into the OFF or high impedance state.

### BLOCK DIAGRAM

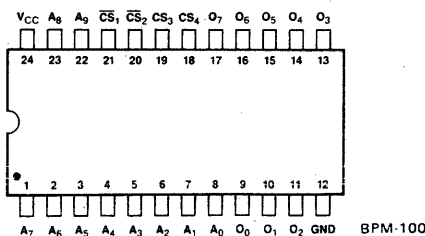


### LOGIC DIAGRAM

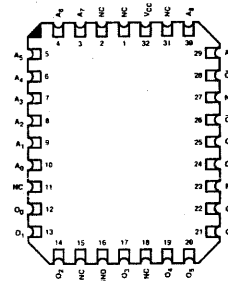


### CONNECTION DIAGRAMS - Top Views

#### DIP



#### Chip-Pak™



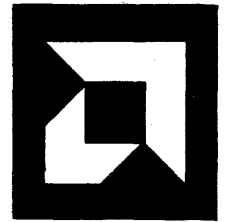
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Advanced Micro Devices

MEMORY

# Advanced Micro Devices



## Bipolar Memories Am27S40A/41A/40/41

### DISTINCTIVE CHARACTERISTICS

- Ultra fast access time "A" version (35ns max) – Fast access time Standard version (50ns max) – allow tremendous system speed improvements
- Platinum-Silicide fuses guarantee high reliability, fast programming and exceptionally high programming yields (typ > 98%)
- AC performance is factory tested utilizing programmed test words and columns
- Voltage and temperature compensated providing extremely flat AC performance over military range
- Member of generic PROM series utilizing standard programming algorithm

### GENERIC SERIES CHARACTERISTICS

These 16K PROMs are members of an Advanced PROM series incorporating common electrical characteristics and programming procedures. All parts in this series are produced with a fusible link at each memory location storing a logic LOW and can be selectively programmed to a logic HIGH by applying appropriate voltages to the circuit.

All parts are fabricated with AMD's fast programming highly reliable Platinum-Silicide Fuse technology. Utilizing easily implemented programming (and common programming personality card sets) these products can be rapidly programmed to any customized pattern. Extra test words are pre-programmed during manufacturing to insure extremely high field programming yields, and produce excellent parametric correlation.

Platinum-Silicide was selected as the fuse link material to achieve a well controlled melt rate resulting in large non-conductive gaps that ensure very stable long-term reliability. Extensive operating testing has proven that this low-field, large-gap technology offers the best reliability for fusible link PROMs.

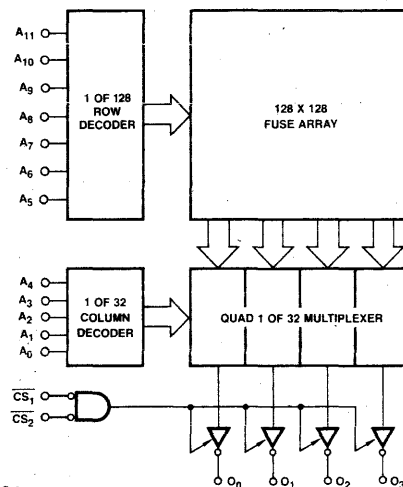
Common design features include active loading of all critical AC paths regulated by a built-in temperature and voltage compensated bias network to provide excellent parametric performance over MIL supply and temperature ranges. Selective feedback techniques have been employed to minimize delays through all critical paths.

These PROMs are manufactured using Advanced Micro Devices' selective oxidation process, IMOX™. This advanced process combined with a merged fuse array permits an increase in density and a decrease in internal capacitance resulting in the fastest possible PROMs.

### FUNCTIONAL DESCRIPTION

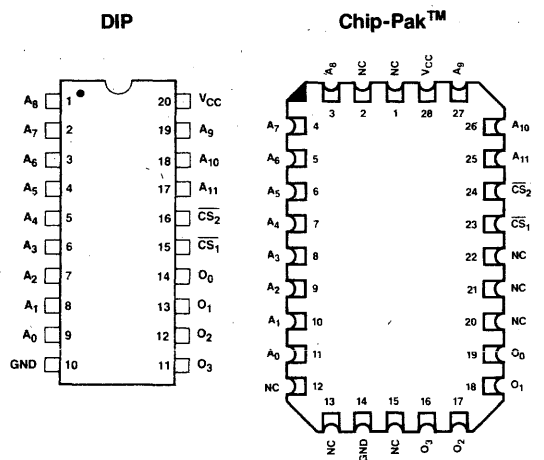
The Am27S40A, Am27S41A, Am27S40, and Am27S41 are high speed electrically programmable Schottky read only memories. Organized in 4096 x 4 configuration, they are available in both open collector (Am27S40A and Am27S40) and three-state (Am27S41A and Am27S41) output versions. After programming, stored information is read on outputs O<sub>0</sub>-O<sub>3</sub> by applying unique binary addresses to A<sub>0</sub>-A<sub>11</sub> and holding the chip select inputs, CS<sub>1</sub> and CS<sub>2</sub>, LOW. If either chip select input goes to a logic HIGH, O<sub>0</sub>-O<sub>3</sub> go to the OFF or HIGH impedance state.

### BLOCK DIAGRAM



BPM-194

### CONNECTION DIAGRAMS – Top Views

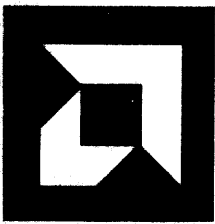


BPM-195

Note: Pin 1 is marked for orientation.

BPM-196

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# Advanced Micro Devices

## Bipolar Memories Am27S43A/43

### DISTINCTIVE CHARACTERISTICS

- Ultra fast access time "A" version (40ns max) – Fast access time Standard version (55ns max) – allow tremendous system speed improvements
- Platinum-Silicide fuses guarantee high reliability, fast programming and exceptionally high programming yields (typ > 98%)
- AC performance is factory tested utilizing programmed test words and columns
- Voltage and temperature compensated providing extremely flat AC performance over military range
- Member of generic PROM series utilizing standard programming algorithm

### GENERIC SERIES CHARACTERISTICS

These 32K PROMs are members of an Advanced PROM series incorporating common electrical characteristics and programming procedures. All parts in this series are produced with a fusible link at each memory location storing a logic LOW and can be selectively programmed to a logic HIGH by applying appropriate voltages to the circuit.

All parts are fabricated with AMD's fast programming highly reliable Platinum-Silicide Fuse technology. Utilizing easily implemented programming (and common programming personality card sets) these products can be rapidly programmed to any customized pattern. Extra test words are pre-programmed during manufacturing to insure extremely high field programming yields, and produce excellent parametric correlation.

Platinum-Silicide was selected as the fuse link material to achieve a well controlled melt rate resulting in large non-conductive gaps that ensure very stable long-term reliability. Extensive operating testing has proven that this low-field, large-gap technology offers the best reliability for fusible link PROMs.

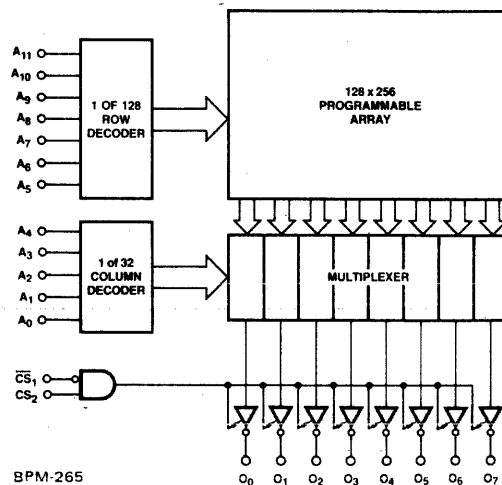
Common design features include active loading of all critical AC paths regulated by a built-in temperature and voltage compensated bias network to provide excellent parametric performance over MIL supply and temperature ranges. Selective feedback techniques have been employed to minimize delays through all critical paths.

These PROMs are manufactured using Advanced Micro Devices' selective oxidation process, IMOX™. This advanced process combined with a merged fuse array permits an increase in density and a decrease in internal capacitance resulting in the fastest possible PROMs.

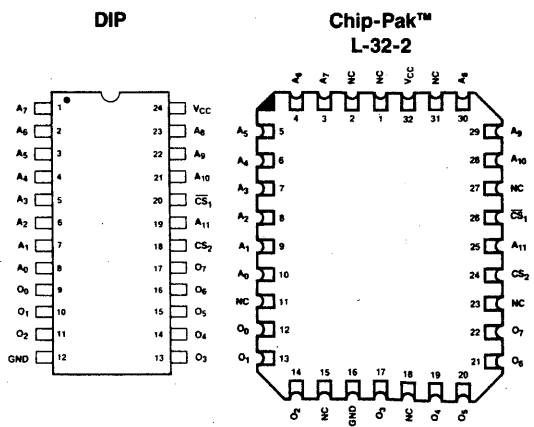
### FUNCTIONAL DESCRIPTION

The Am27S43A and Am27S43 are high speed electrically programmable Schottky read only memories. Organized in 4096 x 8 configuration, they are available in three-state (Am27S43A and Am27S43) output versions. After programming, stored information is read on outputs O<sub>0</sub>-O<sub>7</sub> by applying unique binary addresses to A<sub>0</sub>-A<sub>11</sub> and holding the chip select inputs, CS<sub>1</sub>, LOW and CS<sub>2</sub>, HIGH. If CS<sub>1</sub> goes to logic HIGH or CS<sub>2</sub> goes to a logic LOW, O<sub>0</sub>-O<sub>7</sub> go to the OFF or HIGH impedance state.

### BLOCK DIAGRAM



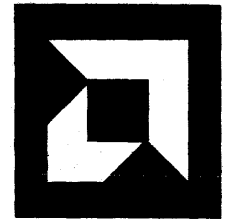
### CONNECTION DIAGRAMS – Top Views



Note: Pin 1 is marked for orientation.

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# Advanced Micro Devices



## Bipolar Memories Am10415SA/415A/415

### DISTINCTIVE CHARACTERISTICS

- Fast access time (10ns typ.) – improves system cycle speeds
- Fully compatible with standard voltage compensated 10K series ECL – no board changes required
- Internally voltage compensated providing flat AC performance
- Outputs preconditioned during write cycle eliminating write recovery glitch
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

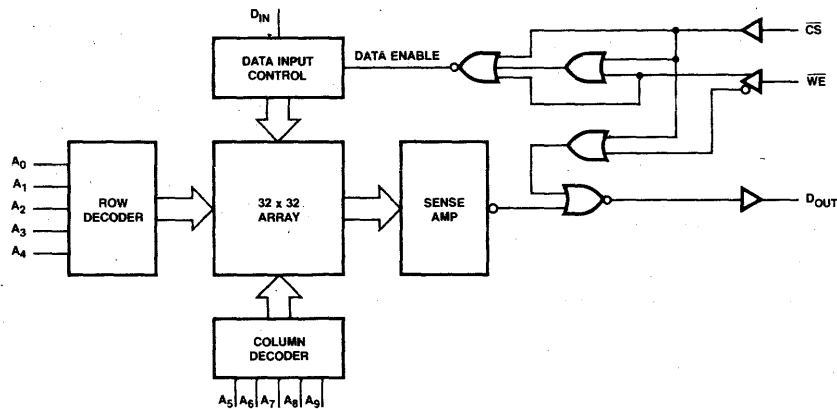
The Am10415SA, Am10415A and Am10415 are fully decoded 1024-bit ECL RAMs organized 1024 words by one bit. Bit selection is achieved by means of a 10-bit address,  $A_0$  through  $A_9$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input ( $D_{IN}$ ) is written into the addressed memory word simultaneously preconditioning the output so true data is present when the write cycle is complete. This preconditioning operation insures minimum write recovery times by eliminating the "write recovery glitch".

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting output ( $D_{OUT}$ ).

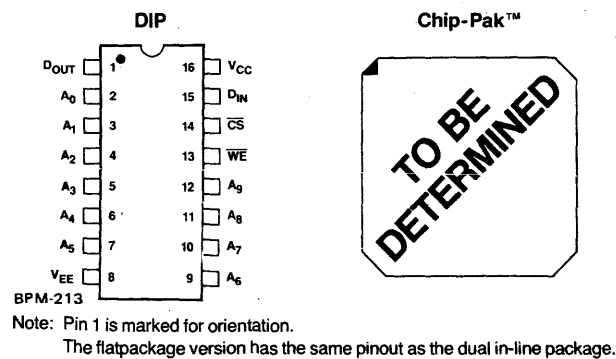
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### BLOCK DIAGRAM



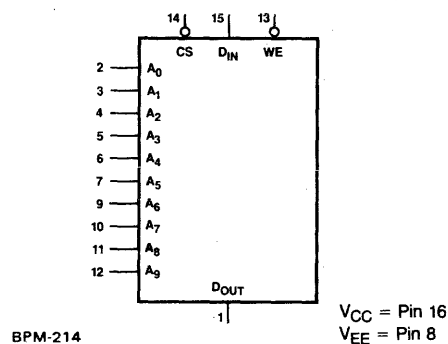
BPM-212

### CONNECTION DIAGRAMS – Top Views



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### LOGIC SYMBOL



BPM-214



# Advanced Micro Devices

## Bipolar Memories Am100415A/415

### DISTINCTIVE CHARACTERISTICS

- Fast access time (10ns typ) – improves system cycle speeds
- Fully compatible with 100K series ECL logic – no board changes required
- Enhanced output voltage level compensation providing 6X (improvement in)  $V_{OL}$  and  $V_{OH}$  stability over supply and temperature ranges
- Internally voltage and temperature compensated providing flat AC performance
- Outputs preconditioned during write cycle eliminating write recovery glitch
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

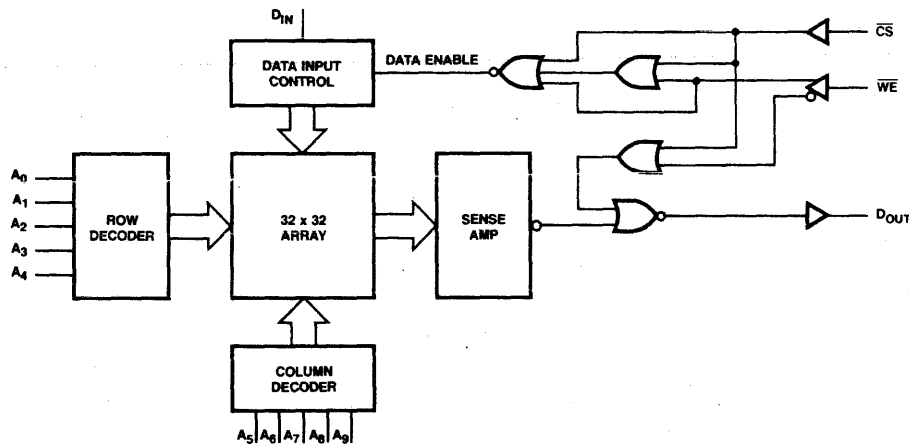
The Am100415A and Am100415 are fully decoded 1024-bit ECL RAMs organized 1024 words by one bit. Bit selection is achieved by means of a 10-bit address,  $A_0$  through  $A_9$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input ( $D_{IN}$ ) is written into the addressed memory word simultaneously preconditioning the output so true data is present when the write cycle is complete. This preconditioning operation insures minimum write recovery times by eliminating the "write recovery glitch".

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting output ( $D_{OUT}$ ).

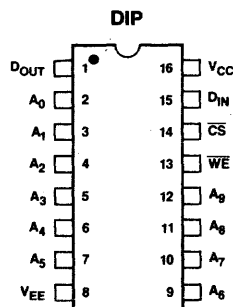
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### LOGIC BLOCK DIAGRAM



BPM-186

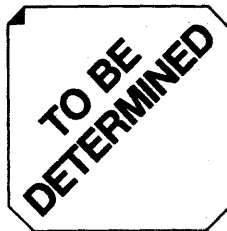
### CONNECTION DIAGRAM – Top Views



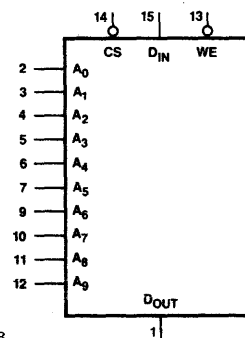
BPM-187

Note: Pin 1 is marked for orientation.

### Chip-Pak™



### LOGIC SYMBOL



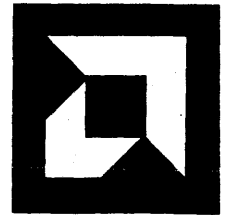
BPM-188

$V_{CC}$  = Pin 16  
 $V_{EE}$  = Pin 8

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# Advanced Micro Devices



## Bipolar Memories Am10474SA/474A/474

### DISTINCTIVE CHARACTERISTICS

- Fast access time (12ns typ) – improves system cycle speeds
- Fully compatible with standard voltage compensated 10K series ECL – no board changes required
- Internally voltage and temperature compensated providing flat AC performance
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

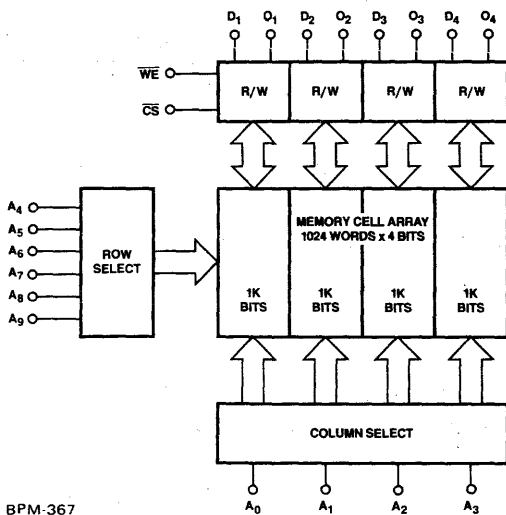
The Am10474SA, Am10474A and Am10474 are fully decoded 4096-bit ECL RAMs organized 1024 words by 4 bits. Word selection is achieved by means of a 10-bit address, A<sub>0</sub> through A<sub>9</sub>. Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input (D<sub>1</sub>-D<sub>4</sub>) are written into the addressed memory words.

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting outputs D<sub>1</sub>-D<sub>4</sub>.

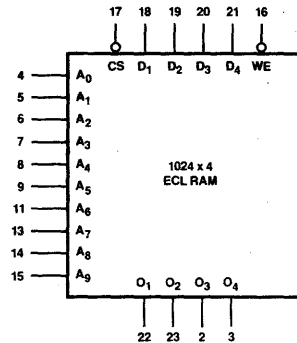
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### LOGIC BLOCK DIAGRAM



BPM-367

### LOGIC SYMBOL



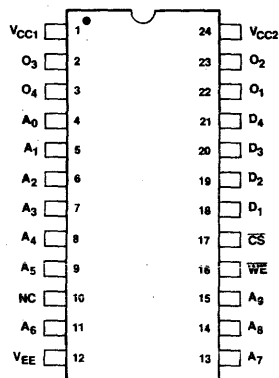
V<sub>CC2</sub> = Pin 24  
V<sub>EE</sub> = Pin 12

BPM-368

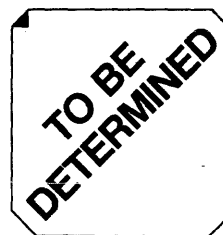
### CONNECTION DIAGRAMS – Top Views

DIP

Chip-Pak™



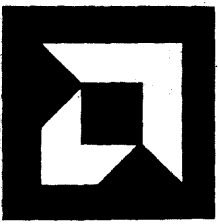
Note: Pin 1 is marked for orientation.



BPM-369

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# Advanced Micro Devices

## Bipolar Memories Am100474SA/474A/474

### DISTINCTIVE CHARACTERISTICS

- Fast access time (12ns typ) – improves system cycle speeds
- Fully compatible with 100K series ECL logic – no board changes required
- Enhanced output voltage level compensation providing 6X (improvement in)  $V_{OL}$  and  $V_{OH}$  stability over supply and temperature ranges
- Internally voltage and temperature compensated providing flat AC performance
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

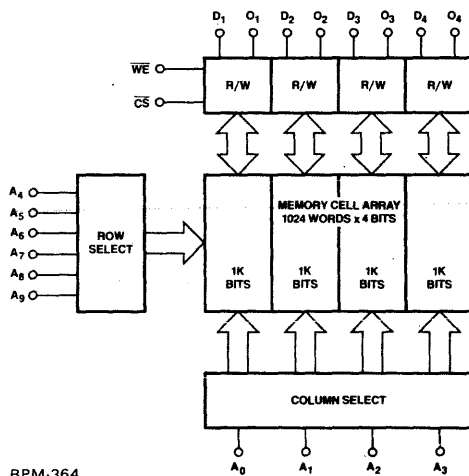
The Am100474SA, Am100474A and Am100474 are fully decoded 4096-bit ECL RAMs organized 1024 words by 4 bits. Word selection is achieved by means of a 10-bit address,  $A_0$  through  $A_9$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and unterminated OR tieable emitter follower outputs.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data inputs ( $D_1$ - $D_4$ ) are written into the addressed memory words.

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed words is read out on the noninverting outputs  $O_1$ - $O_4$ .

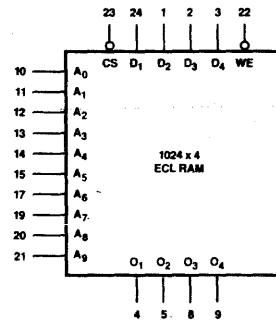
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### LOGIC BLOCK DIAGRAM



BPM-364

### LOGIC SYMBOL

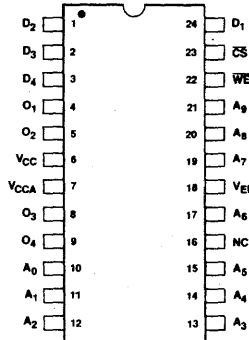


BPM-365

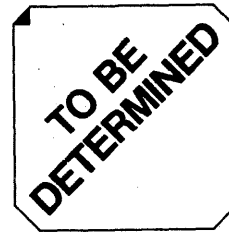
### DIP

### CONNECTION DIAGRAMS – Top Views

### Chip-Pak™



Note: Pin 1 is marked for orientation.



BPM-366

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# Advanced Micro Devices



## Bipolar Memories Am10470SA/470A/470

### DISTINCTIVE CHARACTERISTICS

- Fast access time (12ns typ) – improves system cycle speeds
- Fully compatible with standard voltage compensated 10K series ECL – no board changes required
- Internally voltage compensated providing flat AC performance
- Outputs preconditioned during write cycle eliminating write recovery glitch
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

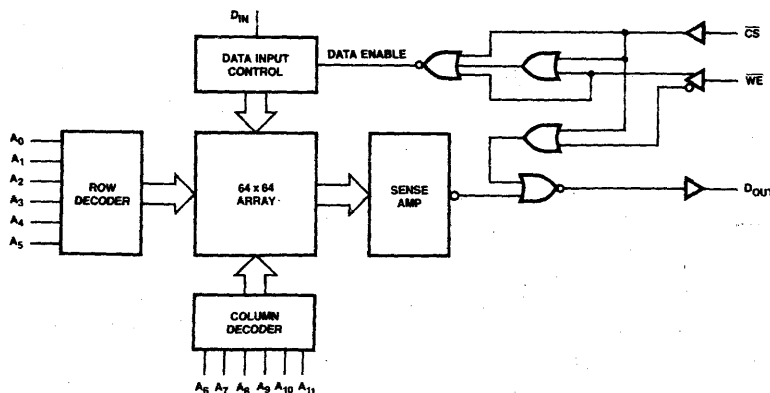
The Am10470SA, Am10470A and Am10470 are fully decoded 4096-bit ECL RAMs organized 4096 words by one bit. Bit selection is achieved by means of a 12-bit address,  $A_0$  through  $A_{11}$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input ( $D_{IN}$ ) is written into the addressed memory word simultaneously preconditioning the output so true data is present when the write cycle is complete. This preconditioning operation insures minimum write recovery times by eliminating the 'write recovery glitch.'

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting output ( $D_{OUT}$ ).

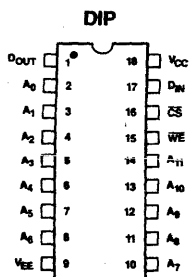
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### BLOCK DIAGRAM



BPM-299

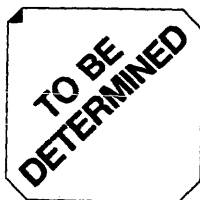
### CONNECTION DIAGRAMS – Top Views



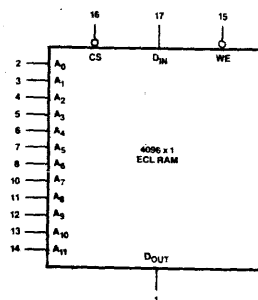
BPM-282

Note: Pin 1 is marked for orientation.

### Chip-Pak™



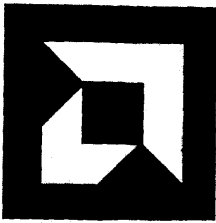
### LOGIC SYMBOL



BPM-283

$V_{CC}$  = Pin 18  
 $V_{EE}$  = Pin 9

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# Advanced Micro Devices

## Bipolar Memories Am100470SA/470A/470

### DISTINCTIVE CHARACTERISTICS

- Fast access time (12ns typ) – improves system cycle speeds
- Fully compatible with 100K series ECL logic – no board changes required
- Enhanced output voltage level compensation providing 6X (improvement in)  $V_{OL}$  and  $V_{OH}$  stability over supply and temperature ranges
- Internally voltage and temperature compensated providing flat AC performance
- Outputs preconditioned during write cycle eliminating write recovery glitch
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

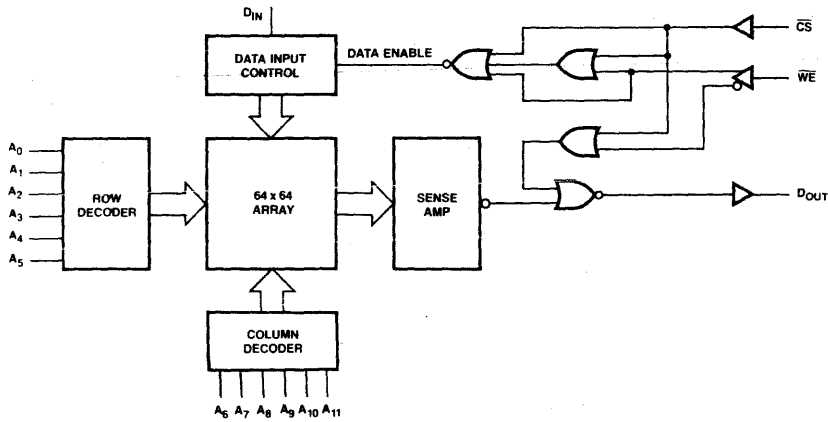
The Am100470SA, Am100470A and Am100470 are fully decoded 4096-bit ECL RAMs organized 4096 words by one bit. Bit selection is achieved by means of a 12-bit address,  $A_0$  through  $A_{11}$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input ( $D_{IN}$ ) is written into the addressed memory word simultaneously preconditioning the output so true data is present when the write cycle is complete. This preconditioning operation insures minimum write recovery times by eliminating the 'write recovery glitch.'

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting output ( $D_{OUT}$ ).

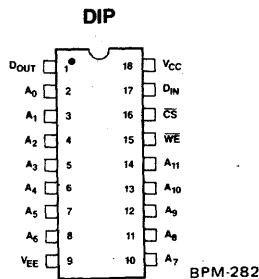
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### LOGIC BLOCK DIAGRAM



BPM-281

### CONNECTION DIAGRAM – Top Views



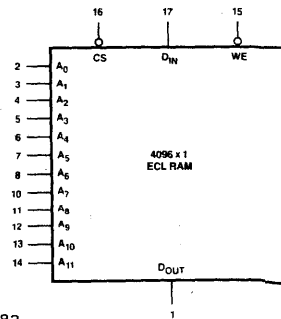
BPM-282

Note: Pin 1 is marked for orientation.

### Chip-Pak™



### LOGIC SYMBOL

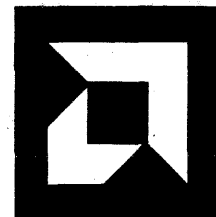


BPM-283

$V_{CC}$  = Pin 18  
 $V_{EE}$  = Pin 9

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# Advanced Micro Devices



## Bipolar Memories Am10470SA/470A/470

### DISTINCTIVE CHARACTERISTICS

- Fast access time (12ns typ) – improves system cycle speeds
- Fully compatible with standard voltage compensated 10K series ECL – no board changes required
- Internally voltage compensated providing flat AC performance
- Outputs preconditioned during write cycle eliminating write recovery glitch
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

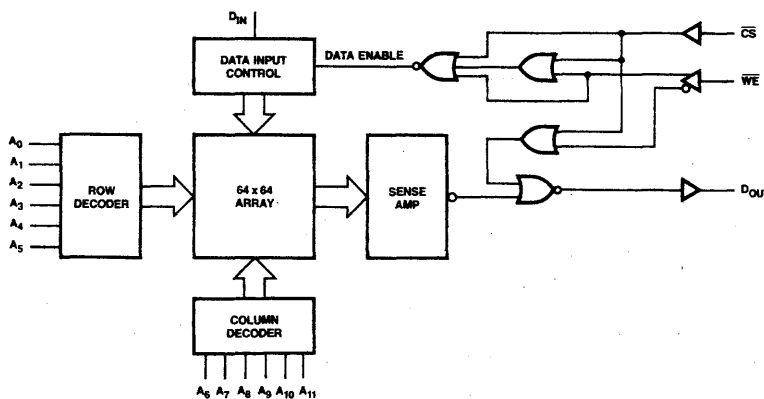
The Am10470SA, Am10470A and Am10470 are fully decoded 4096-bit ECL RAMs organized 4096 words by one bit. Bit selection is achieved by means of a 12-bit address,  $A_0$  through  $A_{11}$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input ( $D_{IN}$ ) is written into the addressed memory word simultaneously preconditioning the output so true data is present when the write cycle is complete. This preconditioning operation insures minimum write recovery times by eliminating the 'write recovery glitch.'

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting output ( $D_{OUT}$ ).

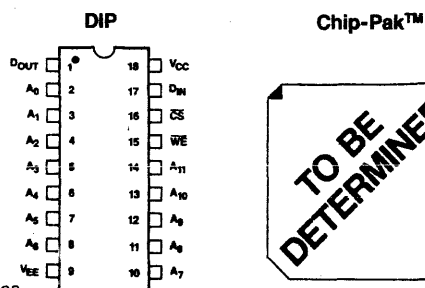
During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

### BLOCK DIAGRAM



BPM-299

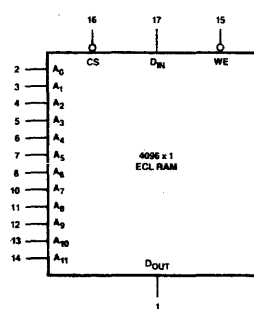
### CONNECTION DIAGRAMS – Top Views



BPM-282

Note: Pin 1 is marked for orientation.

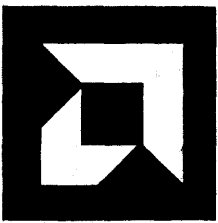
### LOGIC SYMBOL



BPM-283

$V_{CC}$  = Pin 18  
 $V_{EE}$  = Pin 9

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# Advanced Micro Devices

## Bipolar Memories Am100470SA/470A/470

### DISTINCTIVE CHARACTERISTICS

- Fast access time (12ns typ) – improves system cycle speeds
- Fully compatible with 100K series ECL logic – no board changes required
- Enhanced output voltage level compensation providing 6X (improvement in)  $V_{OL}$  and  $V_{OH}$  stability over supply and temperature ranges
- Internally voltage and temperature compensated providing flat AC performance
- Outputs preconditioned during write cycle eliminating write recovery glitch
- Emitter follower outputs – easy wire-ORing
- Power dissipation decreases with increasing temperature

### FUNCTIONAL DESCRIPTION

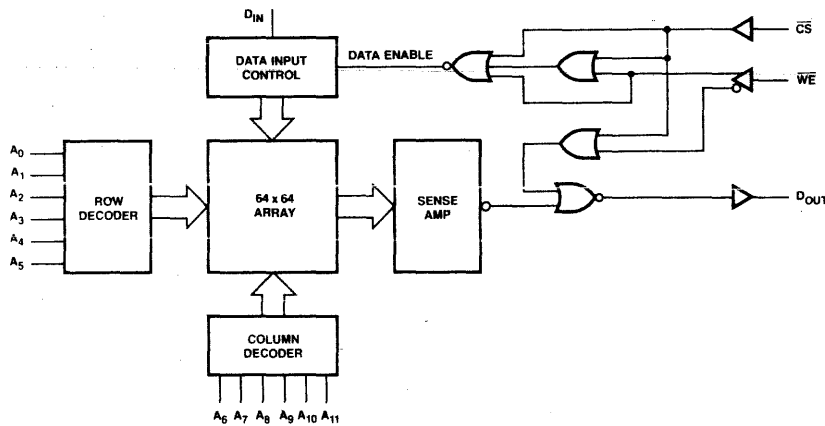
The Am100470SA, Am100470A and Am100470 are fully decoded 4096-bit ECL RAMs organized 4096 words by one bit. Bit selection is achieved by means of a 12-bit address,  $A_0$  through  $A_{11}$ . Easy memory expansion is provided by an active LOW chip select ( $\overline{CS}$ ) input and an unterminated OR tieable emitter follower output.

An active LOW write line ( $\overline{WE}$ ) controls the write/read operation of the memory. When the chip select and write lines are LOW, the data input ( $D_{IN}$ ) is written into the addressed memory word simultaneously preconditioning the output so true data is present when the write cycle is complete. This preconditioning operation insures minimum write recovery times by eliminating the 'write recovery glitch.'

Reading is performed with the chip select line LOW and the write line HIGH. The information stored in the addressed word is read out on the noninverting output ( $D_{OUT}$ ).

During the writing operation or when the chip select line is HIGH the output of the memory goes to a LOW state.

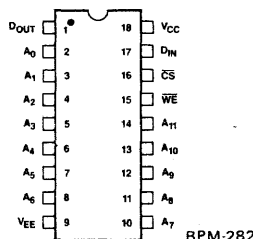
### LOGIC BLOCK DIAGRAM



BPM-281

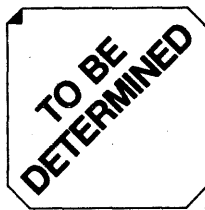
### CONNECTION DIAGRAM – Top Views

#### DIP

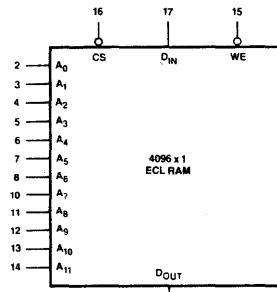


Note: Pin 1 is marked for orientation.

#### Chip-Pak™



### LOGIC SYMBOL

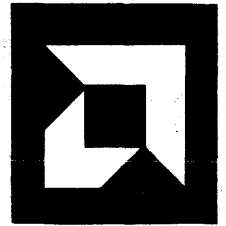


BPM-283

$V_{CC}$  = Pin 18  
 $V_{EE}$  = Pin 9

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# Advanced Micro Devices



## MOS Memories Am2167

### DISTINCTIVE CHARACTERISTICS

- High speed – access times as fast as 35ns maximum
- 16K x 1 organization
- Single +5 volt power supply
- Fully static storage and interface circuitry
- No clocks or timing signals required
- Automatic power down when deselected
- Low power dissipation
  - Am2167: 660mW active, 110mW power down
- Standard 20-pin, .300 inch dual in-line package
- High output drive
  - Up to seven standard TTL loads or six Schottky TTL loads
- TTL compatible interface levels
- No power-on current surge

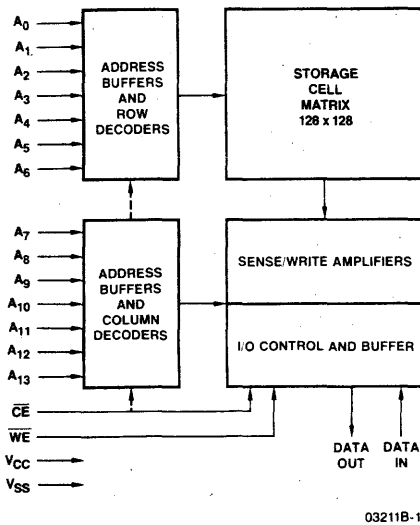
### GENERAL DESCRIPTION

The Am2167 is a high performance, 16,384-bit, static, read/write, random access memory. It is organized as 16,384 words by one bit per word. All interface signal levels are identical to TTL specifications, providing good noise immunity and simplified system design. All inputs are purely capacitive MOS loads. The outputs will drive up to six standard Schottky TTL loads or up to seven standard TTL loads.

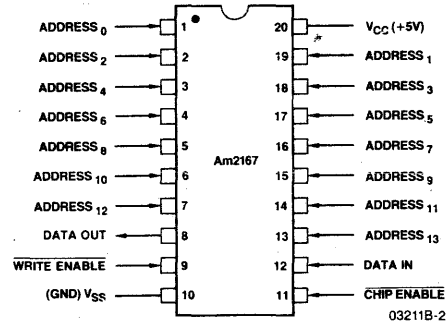
Only a single +5 volt power supply is required. When deselected ( $\overline{CE} \geq V_{IH}$ ), the Am2167 automatically enters a power-down mode which reduces power dissipation by 80%.

Data In and Data Out use separate pins and are the same polarity allowing them to be connected together for operation in a common data bus environment. Data Out is a three-state output allowing similar devices to be wire-OR'd together.

### BLOCK DIAGRAM



### CONNECTION DIAGRAM – Top View D-20-1, P-20-1



Note: Pin 1 is marked for orientation.

### PRODUCT SELECTIONS

	Am2167-35	Am2167-45	Am2167-55	Am2167-70
Maximum Access Time (ns)	35	45	55	70
Maximum Active Current (mA)	120	120 (160 mil)	120 (160 mil)	120 (160 mil)
Maximum Standby Current (mA)	20	20 (30 mil)	20 (30 mil)	20 (30 mil)
Full Military Operating Range Version	No	Yes	Yes	Yes

### ORDERING INFORMATION

Am2167-35 Order Code	Am2167-45 Order Code	Am2167-55 Order Code	Am2167-70 Order Code	Package Type	Operating Range
Am2167-35PC	Am2167-45PC	Am2167-55PC	Am2167-70PC	P-20-1	COM'L
Am2167-35DC	Am2167-45DC	Am2167-55DC	Am2167-70DC	D-20-1	COM'L
–	Am2167-45DM	Am2167-55DM	Am2167-70DM	D-20-1	MIL
–	Am2167-45DMB	Am2167-55DMB	Am2167-70DMB	D-20-1	MIL

Notes: 1. P = Molded DIP, D = Hermetic DIP. Number following letter is number of leads.  
2. See Operating Range table.



# Advanced Micro Devices

## MOS Memories Am9122/L22

### DISTINCTIVE CHARACTERISTICS

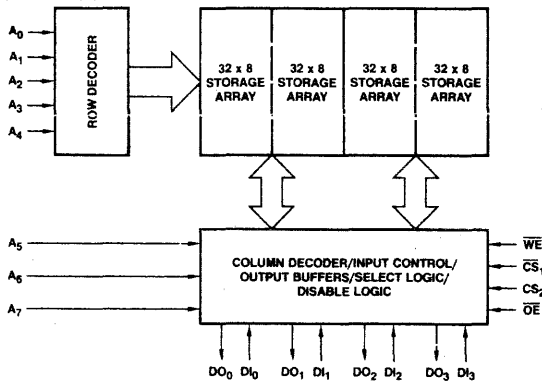
- High performance replacement for 93422/93L422
- 256 x 4 organization for small memory systems
- Fast access times – down to 25ns (Commercial)  
– down to 35ns (Military)
- Low operating power dissipation
  - Standard power: 660mW (Commercial)  
745mW (Military)
  - Low power: 248/440mW (Commercial)  
495mW (Military)
- Single 5 volt power supply – ±10% tolerance both commercial and military
- Guaranteed 0.1% AQL

### FUNCTIONAL DESCRIPTION

The Am9122/Am91L22 series is a MOS pin-for-pin and functional replacement for the 93422/93L422 bipolar memories. These devices are high-performance, low-power, 1024-bit, static, read/write random access memories. They offer a wide range of access times including versions as fast as 25ns. Each memory is implemented as 256 words by 4 bits per word. This organization permits efficient design of small memory systems and allows finer resolution of incremental memory depth.

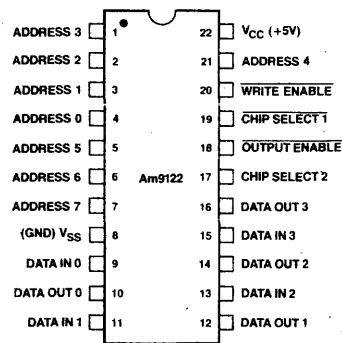
The Am9122/91L22 employs an output enable and two chip enable inputs to give the user better data control. High noise immunity, high output drive (4 TTL loads) and TTL logic voltage levels allow easy conversion from bipolar to MOS. 10% power supply tolerances give better margins in the memory system. As with all AMD MOS RAMs, the Am9122/91L22 is guaranteed to 0.1% AQL.

### Am9122 BLOCK DIAGRAM



RAM-019

### CONNECTION DIAGRAM Top View



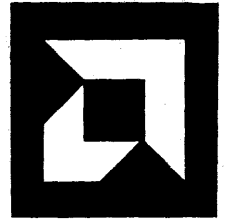
Note: Pin 1 is marked for orientation. RAM-020

### SELECTION GUIDE

		Am9122-25	Am9122-35	Am91L22-35	Am91L22-45	Am91L22-60
Maximum Access Time (ns)		25	35	35	45	60
Maximum Operating Current (mA)	0 to 70°C	120	120	80	80	45
	-55 to 125°C	N/A	135	N/A	90	N/A



# Advanced Micro Devices



## MOS Memories Am9128

### DISTINCTIVE CHARACTERISTICS

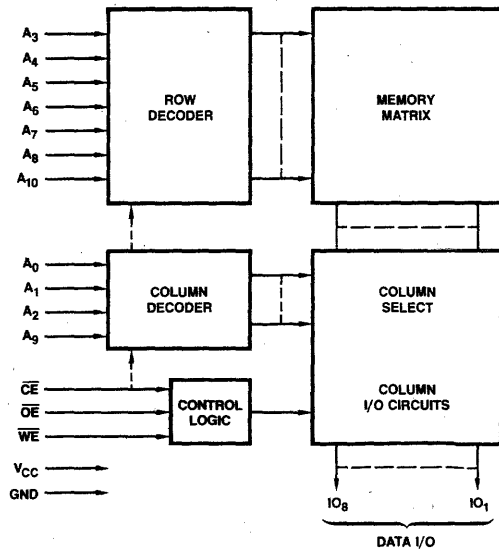
- 2,048 x 8-bit organization
- Logic voltage levels compatible with TTL
- Three-state output buffers-common I/O
- Fully static; no clocks or refresh required
- Single +5V power supply  $\pm 10\%$  tolerance
- $I_{CC}$  max as low as 100mA
- $t_{AA}/t_{ACS}$  as low as 70ns
- Power down mode ( $I_{SB}$  as low as 15mA)
- Commercial and full military temperature ranges

### GENERAL DESCRIPTION

The Am9128 is a 16,384-bit static Random Access Read-write Memory organized as 2048 words of 8 bits. It uses fully static circuitry, requiring no clocks or refresh to operate. Directly TTL-compatible inputs and outputs and operation from a single +5V supply simplify system designs. Common

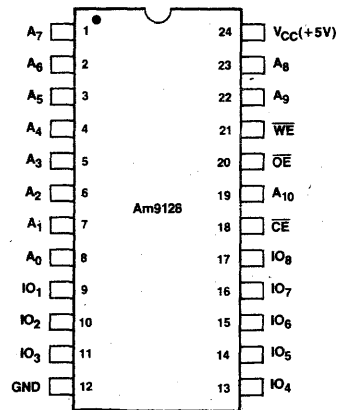
data I/O pins using three-state outputs are provided. The Am9128 is available in an industry-standard 24-pin DIP package with 0.6-inch pin row spacing. The Am9128 uses the JEDEC standard pinout for byte-wide memories (compatible to 16K EPROM's).

### BLOCK DIAGRAM



02050B-1

### CONNECTION DIAGRAM Top View

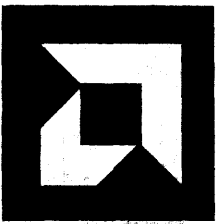


Note: Pin 1 is marked for orientation.

02050B-2

### PRODUCT SELECTOR GUIDE

	Am9128-70	Am9128-90	Am9128-10	Am9128-12	Am9128-15	Am9128-20	
Maximum Access Time (ns)	70	90	100	120	150	200	
0 to 70°C	$I_{CC}$ Max (mA)	140	N/A	120	N/A	100	140
	$I_{SB}$ Max (mA)	30	N/A	15	N/A	15	30
-55 to 125°C	$I_{CC}$ Max (mA)	N/A	180	N/A	150	150	150
	$I_{SB}$ Max (mA)	N/A	30	N/A	30	30	30



# Advanced Micro Devices

## MOS Memories Am9016

### DISTINCTIVE CHARACTERISTICS

- High density 16K x 1 organization
- Replacement for MK4116
- Low maximum power dissipation – 462mW active, 20mW standby
- High-speed operation – 150ns access, 320ns cycle (COM'L)  
200ns access, 375ns cycle (MIL)
- $\pm 10\%$  tolerance on standard +12, +5, -5 voltages
- TTL compatible interface signals
- Three-state output
- RAS only, RMW and Page mode clocking options
- 128 cycle refreshing
- Unlatched data output
- Standard 16-pin, .3 inch wide dual-in-line package
- JEDEC standard 18-pin, Chip-Pak™ leadless chip carrier
- Double poly N-channel silicon gate MOS technology
- Extended ambient operating temperature (-55 to +85°C) available

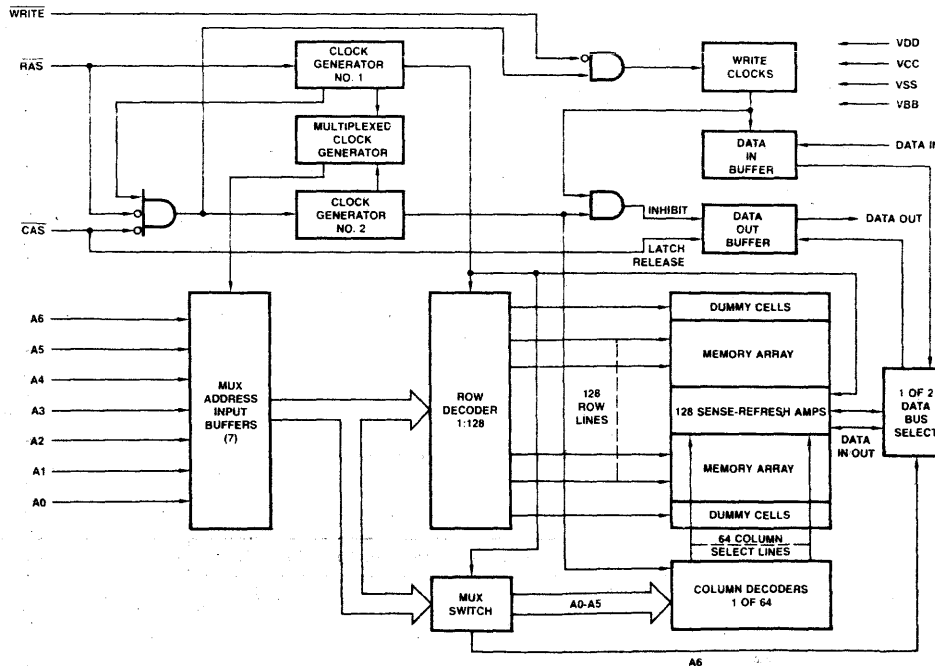
### GENERAL DESCRIPTION

The Am9016 is a high-speed, 16K-bit, dynamic, read/write random access memory. It is organized as 16,384 words by 1 bit per word and is packaged in a standard 16-pin DIP or 18-pin leadless chip carrier. The basic memory element is a single transistor cell that stores charge on a small capacitor. This mechanism requires periodic refreshing of the memory cells to maintain stored information.

All input signals, including the two clocks, are TTL compatible. The Row Address Strobe (RAS) loads the row address and the Column Address Strobe (CAS) loads the column address. The row and column address signals share seven input lines. Active cycles are initiated when RAS goes low, and standby mode is entered when RAS goes high. In addition to normal read and write cycles, other types of operations are available to improve versatility, performance and power dissipation.

The 3-state output buffer turns on when the column access time has elapsed and turns off after CAS goes high. Input and output data are the same polarity.

### BLOCK DIAGRAM



MOS-190

Chip-Pak is a trademark of Advanced Micro Devices, Inc.

# Advanced Micro Devices



## MOS Memories Am9064

### DISTINCTIVE CHARACTERISTICS

- High speed  $\overline{\text{RAS}}$  access of 100 and 120ns
- Fast cycle time of 190 and 220ns
- Single +5V  $\pm$  10% power supply
- Temperature and power supply compensated on-chip substrate-bias generator
- Low power 22mW standby
  - 330mW active – 220ns cycle time
  - 385mW active – 190ns cycle time
- Read, Write, Read-Modify-Write, Page-Mode and  $\overline{\text{RAS}}$ -Only refresh capability
- Early write common I/O capability
- $\overline{\text{CAS}}$  controlled three-state output
- 128 refresh cycles in 2ms
- Double poly NMOS silicon gate technology

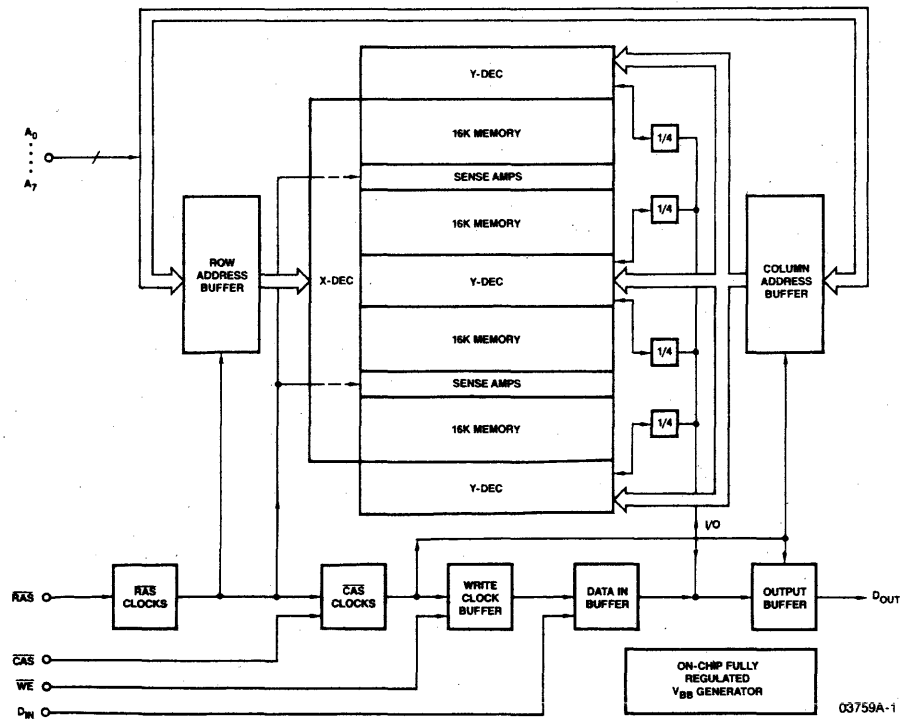
### GENERAL DESCRIPTION

The Am9064 is a high speed, high-performance dynamic RAM, organized 65,536 x 1 and manufactured using advanced NMOS silicon-gate technology. The design is optimized for both high speed and low power dissipation, and only a single +5V supply is needed because the on-chip substrate-bias generator (compensated for temperature and supply variations) provides the necessary back bias.

The Am9064 features multiplexed addressing, and all input signals, including clocks, are TTL-compatible; input and output signals are the same polarity, and the three-state output buffer is  $\overline{\text{CAS}}$  controlled. The Hi-C single-transistor memory cell is used to enhance signal margin and reduce the  $\alpha$ -particle-induced soft-error rate.

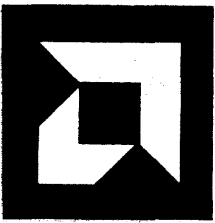
Refresh is achieved by performing  $\overline{\text{RAS}}$ -only cycles or normal read/write cycles on 128 address combinations of ( $A_0 - A_6$ ) within a 2ms period.

### BLOCK DIAGRAM



### ORDERING INFORMATION

Package Type	RAS Access Time		
	100ns	120ns	150ns
Plastic	Am9064-10PC	Am9064-12PC	Am9064-15PC
Cerdip	Am9064-10DC	Am9064-12DC	Am9064-15DC



# Advanced Micro Devices

## MOS Memories Am2168/69

### DISTINCTIVE CHARACTERISTICS

- High speed – access times as fast as 45ns maximum
- 4K x 4 organization
- Single +5 volt power supply
- Fully static storage and interface circuitry
- No clocks or timing signals required
- Automatic power down when deselected
- Power dissipation
  - Am2168: 660mW active; 165mW power down
  - Am2169: 660mW
- Standard 20-pin, .300 inch dual-in-line package
- High output drive
  - Up to seven standard TTL loads or six Schottky TTL loads
- TTL compatible interface levels
- IMOS processing

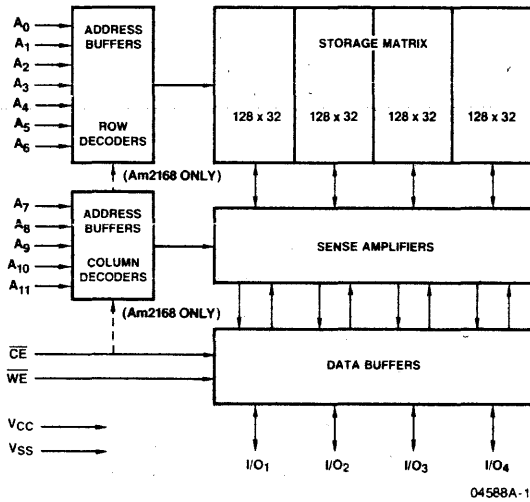
### GENERAL DESCRIPTION

The Am2168 and Am2169 are high performance, static, N-Channel, read/write, random access memories organized as 4096 words of 4 bits. Operation is from a single 5V supply, and all input/output levels are identical to standard TTL specifications. The Am2168 and Am2169 are the same except that the Am2168 offers an automatic  $\overline{CE}$  power down feature.

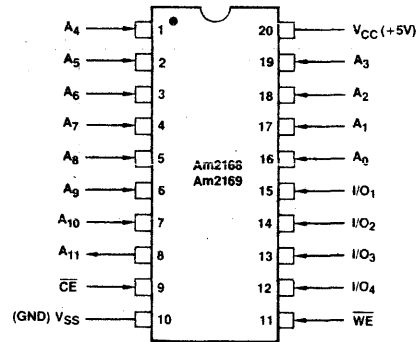
The Am2168 remains in a low-power standby mode as long as  $\overline{CE}$  remains high, thus reducing its power requirements. The Am2168 power decreases from 660mW to 165mW in the standby mode. The  $\overline{CE}$  input does not affect the power dissipation of the Am2169.

Data readout is not destructive and has the same polarity as data input.  $\overline{CE}$  provides for easy selection of an individual device when the outputs are OR-tied.

### BLOCK DIAGRAM



### CONNECTION DIAGRAM Top View



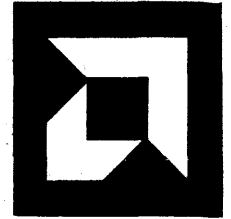
Note: Pin 1 is marked for orientation.

### PRODUCT SELECTIONS

		Am2168-45	Am2169-45	Am2168-55	Am2169-55	Am2168-70	Am2169-70
Maximum Access Time (ns)		45	45	55	55	70	70
0 to 70°C	$I_{CC}$ (mA)	120	120	120	120	120	120
	$I_{SB}^*$ (mA)	30	N/A	30	N/A	30	N/A
-55 to +125°C	$I_{CC}$ (mA)	N/A	N/A	160	160	160	160
	$I_{SB}^*$ (mA)	N/A	N/A	30	30	30	30

\*Am2168 only

# Advanced Micro Devices



## MOS Memories Am9150

### DISTINCTIVE CHARACTERISTICS

- 1024 x 4 organization
- High speed – 25ns max access time
- Separate data inputs and outputs
- Memory reset function
- High density SLIM 24-pin 300-MIL package
- Three-state output buffers
- Single +5V power supply  $\pm 10\%$

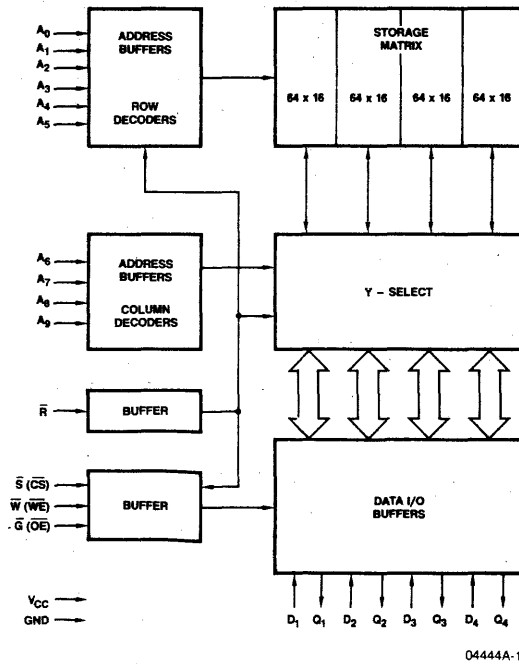
### GENERAL DESCRIPTION

The Am9150 is a high-performance, static, n-channel, read/write, random access memory organized as 1024 x 4. It features single 5V supply operation, TTL-compatible input and output levels, and separate input and output pins for improved system performance and ease of use.

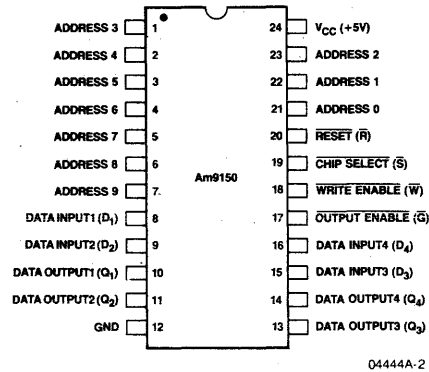
The Am9150 also incorporates a reset feature which will reset the entire contents of the memory to a logical LOW in two cycle times by controlling  $\bar{R}$  (RESET) and  $\bar{S}$  ( $\bar{CS}$ ).

The Am9150 has four control signals,  $\bar{R}$ ,  $\bar{S}$ ,  $\bar{W}$  and  $\bar{G}$ . The  $\bar{S}$  input controls read, write and reset operations of the device and provides for easy selection of an individual device when the outputs are tied together. The  $\bar{W}$  (WE) input controls the normal read and write operations, and the  $\bar{G}$  ( $\bar{OE}$ ) controls the state of the outputs.

### BLOCK DIAGRAM



### CONNECTION DIAGRAM Top View



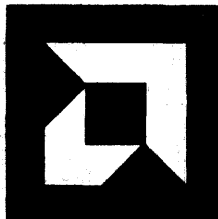
Note: Pin 1 is marked for orientation.

### FUNCTION TABLE

$\bar{S}$	Inputs			Outputs	Mode
	$\bar{W}$	$\bar{G}$	$\bar{R}$		
H	X	X	X	High Z	Not Selected
L	H	X	L	High Z	Reset
L	L	X	H	High Z	Write
L	H	L	H	$Q_1 - Q_4$	Read
L	X	H	H	High Z	Output Disable

### PRODUCT SELECTIONS

		Am9150-25	Am9150-35	Am9150-45
Maximum Access Time (ns)		25	35	45
$I_{CC}$ Max (mA)	0 to +70°C	180	180	180
	-55 to +125°C	N/A	180	180



# Advanced Micro Devices

## MOS Memories Am9264

### DISTINCTIVE CHARACTERISTICS

- Enhanced manufacturability with post-metal programming
- Access time – 250ns (max)
- Single +5V ±10% power supply
- Fully static operation
- Completely TTL compatible
- Standard 24 pin DIP
- Pin compatible with 16K/32K/64K EPROMs/ROMs
- Military version (–55 to +125°C) – Available – 450ns (max) access time

### FUNCTIONAL DESCRIPTION

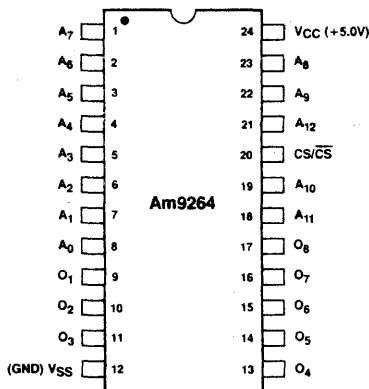
The Am9264 high performance read only memory is organized 8192 words by 8 bits with access times of less than 250ns. This organization simplifies the design of small memory systems and permits incremental memory sizes of 8192 words. The fast access times provided allow the ROM to service high performance microcomputer applications without stalling the processor.

The programmable chip select input signal is provided to control the output buffers. Chip Select Polarity may be provided by the customer thus allowing the addressing of 2 memory chips without external gating. The outputs of the unselected chips are turned off and assume a high impedance state. This permits wire-ORing with additional Am9264 devices and other three state components.

This memory is fully static and requires no clock signals of any kind. A selected chip will output data from a location specified by the address present on the address input lines. Input and output levels are compatible with TTL specifications.

The ability to program customer code at the last step of fabrication (Post Metal Programming Technique) will result in faster turn around time for new or old patterns. This technique will allow us to test wafers before committing customer patterns to categorize speed and power dissipation requirements.

### CONNECTION DIAGRAM Top View



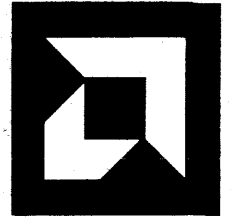
Note: Pin 1 is marked for orientation.

ROM-010

### ORDERING INFORMATION

Package Type	Ambient Temperature Specifications	Access Time		
		450ns	300ns	250ns
Molded	0°C ≤ T <sub>A</sub> ≤ +70°C	AM9264BPC	AM9264CPC	AM9264DPC
Cerdip	0°C ≤ T <sub>A</sub> ≤ +70°C	AM9264BCC	AM9264CCC	AM9264DCC
Ceramic Side-Brazed	0°C ≤ T <sub>A</sub> ≤ +70°C	AM9264BDC	AM9264CDC	AM9264DDC
	–55°C ≤ T <sub>A</sub> ≤ +125°C	AM9264BDM		

# Advanced Micro Devices



## MOS Memories Am9265

### DISTINCTIVE CHARACTERISTICS

- Enhanced manufacturability with post metal programming
- Access time – 250ns (max)
- Fully static operation
- Single +5V ± 10% power supply
- Automatic power down feature controlled by separate  $\overline{CE}$  pin.
  - 80mA max operating current
  - 20mA max standby current
- Separate  $\overline{OE}$  pin for tri-state output control
- Two programmable chip selects with no-connect option
- Pin compatible with 28 pin 64K and higher density ROMs/EPROMs
- Completely TTL compatible
- Standard 28 pin DIP
- Military version (–55 to +125°C) – Available with 450ns (max) access time

### FUNCTIONAL DESCRIPTION

The Am9265 high performance read only memory is organized 8192 words by 8 bits with access times of less than 250ns. This organization simplifies the design of small memory systems and permits incremental memory sizes of 8192 words. The fast access times provided allow the ROM to service high performance microcomputer applications without stalling the processor.

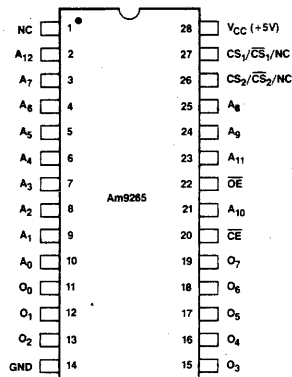
Two programmable chip select inputs are provided to control the output buffers. Chip select polarity may be specified by the customer thus allowing the addressing of 4 memory chips without external gating. The outputs of the unselected chips are turned off and assume a high impedance state. This permits wire-ORing with additional Am9265 devices and other three state components. No-connect option on chip selects can be provided if desired by the customer.

This memory is fully static and requires no clock signals of any kind. A selected chip will output data from a location specified by the address present on the address input lines. Input and output levels are compatible with TTL specifications. A separate OE, output enable pin, controls outputs providing greater system flexibility and eliminating bus contention.

The Am9265 features an automatic stand-by mode. When deselected by  $\overline{CE}$ , the maximum supply current is reduced from 80mA to 20mA, a 75% reduction.

The ability to program customer code at the last step of fabrication (Post Metal Programming Technique) results in faster turn around time for new or old patterns. This technique also allows testing of wafers to categorize speed and power dissipation before committing customer patterns.

### CONNECTION DIAGRAM Top View



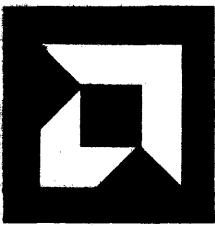
Note: Pin 1 is marked for orientation ROM-001

### OPERATING RANGE

Part Number	Ambient Temperature	V <sub>CC</sub>	V <sub>SS</sub>
Am9265DC/PC/CC	0°C ≤ T <sub>A</sub> ≤ +70°C	+5.0V ± 10%	0V
Am9265DM	–55°C ≤ T <sub>A</sub> ≤ +125°C	+5.0V ± 10%	0V

### ORDERING INFORMATION

Package Type	Ambient Temperature Specifications	Access Time		
		450ns	300ns	250ns
Molded	0°C ≤ T <sub>A</sub> ≤ +70°C	AM9265BPC	AM9265CPC	AM9265DPC
Cerdip	0°C ≤ T <sub>A</sub> ≤ +70°C	AM9265BCC	AM9265CCC	AM9265DCC
Ceramic Side-Brazed	0°C ≤ T <sub>A</sub> ≤ +70°C	AM9265BDC	AM9265CDC	AM9265DDC
	–55°C ≤ T <sub>A</sub> ≤ +125°C	AM9265BDM		



# Advanced Micro Devices

## MOS Memories Am92128

### DISTINCTIVE CHARACTERISTICS

- Enhanced manufacturability with post metal programming
- Access time – 250ns (max)
- Fully static operation
- Single +5V ± 10% power supply
- Automatic power down feature controlled by separate  $\overline{CE}$  pin
  - 80mA max operating current
  - 25mA max standby current
- Separate  $\overline{OE}$  pin for three-state output control
- Programmable chip select with no-connect option
- Pin compatible with 28-pin and high density ROMs/EPROMs
- TTL compatible
- Standard 28-pin DIP
- Military version (–55 to +125°C) – Available with 450ns (max) access time

### FUNCTIONAL DESCRIPTION

The Am92128 high performance read only memory is organized 16,384 words by 8 bits and has access times of less than 250ns. This organization simplifies the design of memory systems and permits incremental memory sizes of 16,384 bytes. The fast access times provided allow the ROM to service high performance microcomputer applications without inserting wait states.

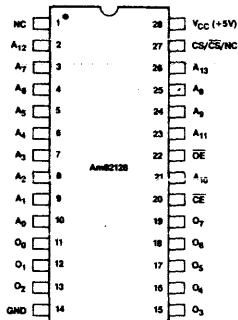
One programmable chip select input is provided to control the output buffers. Chip select polarity may be specified by the customer thus allowing the addressing of 2 memory chips without external gating. The outputs of the unselected chips are turned off and assume a high impedance state. This permits wire-ORing with additional Am92128 devices and other three-state components. No-connect option on chip select can be provided if desired by the customer.

This memory is fully static and requires no clock signals of any kind. A selected chip will output data from a location specified by the address present on the address input lines. Input and output levels are compatible with TTL specifications. A separate  $\overline{OE}$ , output enable pin, controls outputs providing greater system flexibility and eliminating bus contention.

The Am92128 features an automatic stand-by mode. When deselected by  $\overline{CE}$ , the maximum supply current is reduced from 80mA to 25mA, a 70% reduction.

The ability to program customer code at the last step of fabrication (Post Metal Programming Technique) results in faster turn around time for new or old patterns. This technique also allows testing of wafers to categorize speed and power dissipation before committing customer patterns.

### CONNECTION DIAGRAM Top View



Note: Pin 1 is marked for orientation.

ROM-008

### PIN NAMES

$A_0 - A_{13}$	Address	$\overline{OE}$	Output Enable
$\overline{CE}$	Chip Enable	$V_{CC}$	+5V
NC	No Connection	GND	Ground
CS/ $\overline{CS}$	Chip Select	$Q_0 - Q_7$	Data Outputs

### TRUTH TABLE

$\overline{CS}$ or CS	$\overline{CE}$	$\overline{OE}$	Mode	Outputs	Power	
H	L	L	X	Deselected	High-Z	Active
H	L	H	X	Deselected	High-Z	Standby
L	H	L	H	Inhibit	High-Z	Active
L	H	H	X	Deselected	High-Z	Standby
L	H	L	L	Read	$D_{OUT}$	Active

H = HIGH ( $\geq 2.0V$ )

L = LOW ( $\leq 0.8V$ )

X = Don't Care

### OPERATING RANGE

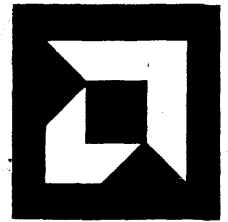
Part Number	Ambient Temperature		$V_{CC}$	$V_{SS}$
	Specifications			
Am92128 X PC/CC	$0^\circ C \leq T_A \leq +70^\circ C$		+5.0V ± 10%	0V
Am92128BDM	$-55^\circ C \leq T_A \leq +125^\circ C$		+5.0V ± 10%	0V

### ORDERING INFORMATION

Package Type	Ambient Temperature Specifications	Access Time		
		450ns	300ns	250ns
Molded	$0^\circ C \leq T_A \leq +70^\circ C$	Am92128BPC	Am92128CPC	Am92128DPC
Cerdip	$0^\circ C \leq T_A \leq +70^\circ C$	Am92128BCC	Am92128CCC	Am92128DCC
Side-Brazed	$+55^\circ C \leq T_A \leq +125^\circ C$	Am92128BDM		



# Advanced Micro Devices



## MOS Memories Am92256

### DISTINCTIVE CHARACTERISTICS

- Enhanced manufacturability with post metal programming
- Access time – 250ns (max)
- Fully static operation
- Single +5V  $\pm$  10% power supply
- Automatic power down feature controlled by separate  $\overline{CE}$  pin
  - 120mA max operating current
  - 30mA max standby current
- Separate  $\overline{OE}$  pin for three-state output control
- Pin compatible with 28-pin high density ROMs/EPROMs
- TTL compatible
- Standard 28-pin DIP

### FUNCTIONAL DESCRIPTION

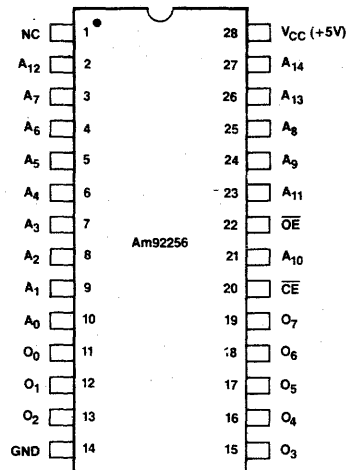
The Am92256 high performance read only memory is organized 32,768 words by 8 bits and has access times of less than 250ns. This organization simplifies the design of memory systems and permits incremental memory sizes of 32,768 bytes. The fast access times provided allow the ROM to service high performance microcomputer applications without inserting wait states.

The Am92256 features an automatic stand-by mode. When deselected by  $\overline{CE}$ , the maximum supply current is reduced from 120mA to 30mA, a 75% reduction. The outputs of the deselected chips are turned off and assume a high impedance state. This permits wire-ORing with additional Am92256 devices and other three-state components.

This memory is fully static and requires no clock signals of any kind. A selected chip will output data from a location specified by the address present on the address input lines. Input and output levels are compatible with TTL specifications. A separate  $\overline{OE}$ , output enable pin, controls outputs providing greater system flexibility and eliminating bus contention.

The ability to program customer code at the last step of fabrication (Post Metal Programming Technique) results in faster turn around time for new or old patterns. This technique also allows testing of wafers to categorize speed and power dissipation before committing customer patterns.

### CONNECTION DIAGRAM Top View



Note: Pin 1 is marked for orientation. ROM-005

### PIN NAMES

A <sub>0</sub> –A <sub>14</sub>	Address	$\overline{OE}$	Output Enable
$\overline{CE}$	Chip Enable	V <sub>CC</sub>	+5V
NC	No Connection	GND	Ground
		O <sub>0</sub> –O <sub>7</sub>	Data Outputs

### TRUTH TABLE

$\overline{CE}$	$\overline{OE}$	Mode	Outputs	Power
H	X	Deselect	High-Z	Standby
L	H	Inhibit	High-Z	Active
L	L	Read	D <sub>OUT</sub>	Active

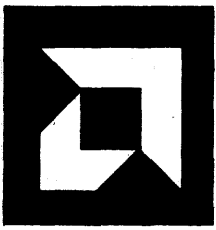
H = HIGH ( $\geq 2.0V$ )  
L = LOW ( $\leq 0.8V$ )  
X = Don't Care

### OPERATING RANGE

Part Number	Ambient Temperature	V <sub>CC</sub>	V <sub>SS</sub>
Am92256PC/CC	0°C $\leq$ T <sub>A</sub> $\leq$ +70°C	+5.0V $\pm$ 10%	0V

### ORDERING INFORMATION

Package Type	Ambient Temperature Specifications	Access Time		
		450ns	300ns	250ns
Molded	0°C $\leq$ T <sub>A</sub> $\leq$ +70°C	Am92256BPC	Am92256CPC	Am92256DPC
Cerdip	0°C $\leq$ T <sub>A</sub> $\leq$ +70°C	Am92256BCC	Am92256CCC	Am92256DCC



# Advanced Micro Devices

## MOS Memories Am2732

### DISTINCTIVE CHARACTERISTICS

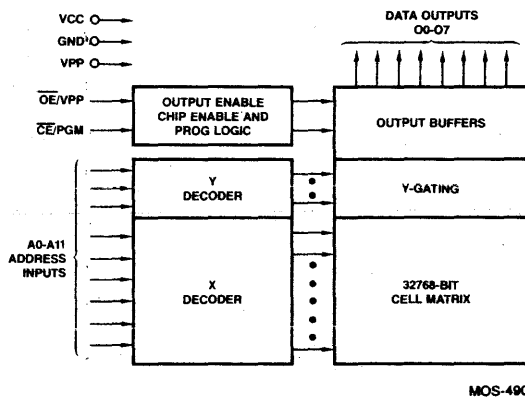
- 0.3% AQL guaranteed
- Direct replacement for Intel 2732
- Pin compatible with Am9233 – 32K ROM
- Single +5V power supply
- Fast access time – 350ns and 450ns
- Low power dissipation
  - 787mW active
  - 157mW standby
- Fully static operation – no clocks
- Three-state outputs
- TTL compatible inputs/outputs

### GENERAL DESCRIPTION

The Am2732 is a 32768-bit ultraviolet erasable and programmable read-only memory. It is organized as 4096 words by 8 bits per word, operates from a single +5V supply, has a static standby mode, and features fast single address location programming.

Because the Am2732 operates from a single +5V supply, it is ideal for use in microprocessor systems. All programming signals are TTL levels, requiring a single pulse. For programming outside of the system, existing EPROM programmers may be used. Locations may be programmed singly, in blocks, or at random. Total programming time for all bits is 200 seconds.

### BLOCK DIAGRAM

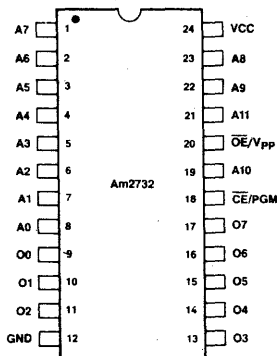


### MODE SELECTION

Pins	$\overline{CE}/PGM$ (18)	$\overline{OE}/VPP$ (20)	VCC (24)	Outputs (9-11, 13-17)
Read	VIL	VIL	+5	DOUT
Standby	VIH	Don't Care	+5	High Z
Program	VIL	VPP	+5	DIN
Program Verify	VIL	VIL	+5	DOUT
Program Inhibit	VIH	VPP	+5	High Z

### CONNECTION DIAGRAMS – Top Views

#### DIP

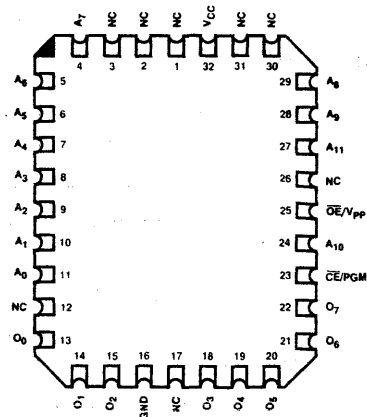


Pin 1 is marked for orientation.

A<sub>0</sub>-A<sub>11</sub>: Addresses  
 O<sub>0</sub>-O<sub>7</sub>: Outputs  
 $\overline{CE}/PGM$ : Chip Enable/Program  
 $\overline{OE}/VPP$ : Output Enable

MOS-491

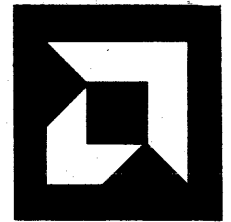
#### Chip-Pak™ L-32-2



MOS-671

Chip-Pak is a trademark of Advanced Micro Devices, Inc.

# Advanced Micro Devices



## MOS Memories Am2732A

### DISTINCTIVE CHARACTERISTICS

- Fast access times – 200ns, 250ns, 300ns, 450ns
- New low-cost plastic package for applications not requiring reprogramming
- Low power dissipation
  - 525mW active, 130mW standby
- $\pm 10\%$   $V_{CC}$  supply tolerance available
- Three-state outputs
- 24-pin JEDEC approved 2732 pin-out
- Pin compatible with Am9233 – 32K-bit ROM
- Separate chip enable and output enable

### GENERAL DESCRIPTION

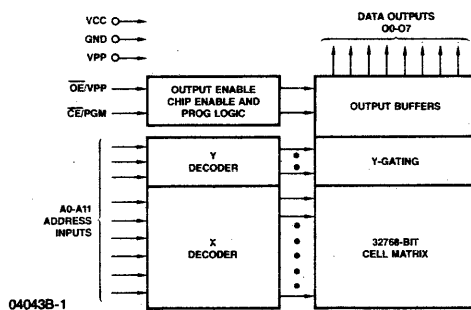
The Am2732A is a 32768-bit UV-light erasable and electrically programmable read-only memory, organized as 4096 words by 8-bits. The standard Am2732A offers an access time of 250ns, allowing operation with high-speed microprocessors without any WAIT state.

To eliminate bus contention in a multiple-bus microprocessor system, Am2732A offers separate Output Enable (OE) and Chip Enable (CE) controls.

All signals are TTL levels, including programming signals. Bit locations may be programmed singly, in blocks or at random.

The part is available in an economical plastic package for applications which do not require reprogramming.

### BLOCK DIAGRAM



04043B-1

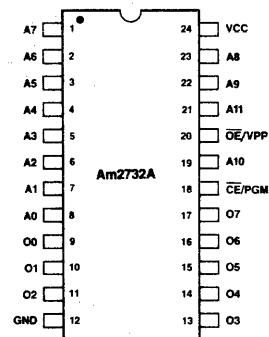
### MODE SELECTION

Mode	Pins CE/PGM (16)	OE/VPP (20)	VCC (24)	Outputs (9-11, 13-17)
Read	VIL	VIL	+5	DOUT
Standby	VIH	X	+5	High Z
Program	VIL	VPP	+5	DIN
Program Verify	VIL	VIL	+5	DOUT
Program Inhibit	VIH	VPP	+5	High Z

X can be either V<sub>IL</sub> or V<sub>HI</sub>

Figure 1.

### CONNECTION DIAGRAM – Top View D-24-4



04043B-2

Pin 1 is marked for orientation.

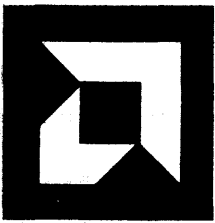
A0-A11: Addresses  
O0-O7: Outputs  
CE/PGM: Chip Enable/Program  
OE/VPP: Output Enable/VPP Supply

Figure 2.

### ORDERING INFORMATION

Order Number <sup>1</sup>	t <sub>ACC</sub> (ns)	t <sub>CE</sub> (ns)	t <sub>OE</sub> (ns)	V <sub>CC</sub>	Temperature Range
AM2732APC	250	250	100	5V $\pm 5\%$	C
AM2732A-2DX	200	200	70	5V $\pm 5\%$	C, I, L
AM2732A-20DX	200	200	70	5V $\pm 10\%$	C, I, L, M
AM2732A-DX	250	250	100	5V $\pm 5\%$	C, I, L
AM2732A-25DX	250	250	100	5V $\pm 10\%$	C, I, L, M
AM2732A-3DX	300	300	150	5V $\pm 5\%$	C, I, L
AM2732A-30DX	300	300	150	5V $\pm 10\%$	C, I, L
AM2732A-4DX	450	450	150	5V $\pm 5\%$	C, I, L
AM2732A-45DX	450	450	150	5V $\pm 10\%$	C, I, L, M

Note 1: X = C (0 to 70°C), X = I (-40 to +85°C), X = L (-55 to +100°C), X = M (-55 to +125°C).



# Advanced Micro Devices

## MOS Memories Am2764

### DISTINCTIVE CHARACTERISTICS

- Fast access time – 200ns, 250ns, and 300ns
- New low-cost plastic package for applications not requiring reprogramming
- Low power dissipation
  - 525mW active, 105mW standby
- $\pm 10\%$  power supply tolerance available
- 28-pin JEDEC approved 2764 pinout
- Pin compatible with Am9265 – 64K ROM
- Choice of 4 temperature ranges in ceramic package
- Fast programming time

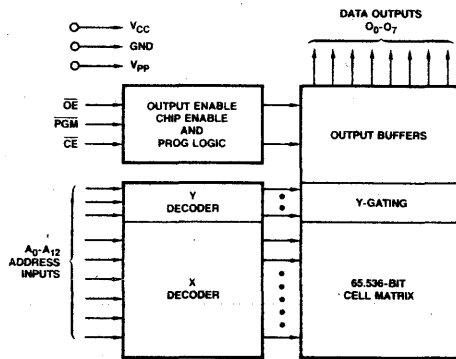
### GENERAL DESCRIPTION

The Am2764 is a 65536-bit ultraviolet erasable and programmable read-only memory. It is organized as 8192 words by 8 bits per word, operates from a single +5V supply, has a static standby mode, and features fast single address location programming.

Because the Am2764 operates from a single +5V supply, it is ideal for use in microprocessor systems. All programming signals are TTL levels, requiring a single pulse. For programming outside of the system, existing EPROM programmers may be used. Locations may be programmed singly, in blocks, or at random.

The part is available in an economical plastic package for applications which do not require reprogramming.

### BLOCK DIAGRAM



01942D-1

### MODE SELECTION

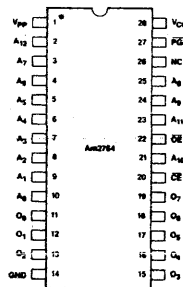
Mode	Pins	CE (20)	OE (22)	PGM (27)	Vpp (1)	VCC (28)	Outputs (11-13, 15-19)
Read		V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>CC</sub>	V <sub>CC</sub>	DOUT
Standby		V <sub>IH</sub>	X	X	V <sub>CC</sub>	V <sub>CC</sub>	High Z
Program		V <sub>IL</sub>	X	V <sub>IL</sub>	V <sub>PP</sub>	V <sub>CC</sub>	DIN
Program Verify		V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>PP</sub>	V <sub>CC</sub>	DOUT
Program Inhibit		V <sub>IH</sub>	X	X	V <sub>PP</sub>	V <sub>CC</sub>	High Z

X can be either V<sub>IL</sub> or V<sub>IH</sub>

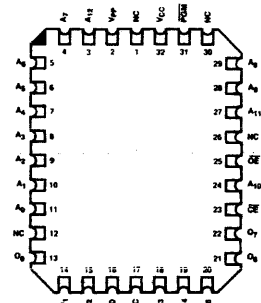
### CONNECTION DIAGRAMS – Top Views

D-28-1  
P-28-1

L-32-2



01942D-2



01942D-3

Pin 1 is marked for orientation.

A<sub>0</sub>-A<sub>12</sub>: Addresses  
O<sub>0</sub>-O<sub>7</sub>: Outputs  
CE: Chip Enable

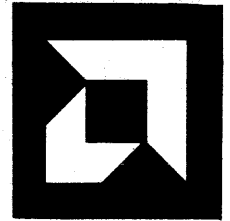
OE: Output Enable  
PGM: Program  
NC: No Connect

### ORDERING INFORMATION

Order Number <sup>1</sup>	t <sub>ACC</sub> (ns)	t <sub>CE</sub> (ns)	t <sub>OE</sub> (ns)	V <sub>CC</sub>	Temperature Range
AM2764-2DX AM2764-2LX	200	200	75	5V $\pm$ 5%	C, I
AM2764-20DX AM2764-20LX	200	200	75	5V $\pm$ 10%	C, I
AM2764DX AM2764LX	250	250	100	5V $\pm$ 5%	C, I, L
AM2764-25DX AM2764-25LX	250	250	100	5V $\pm$ 10%	C, I, L, M
AM2764-3DX AM2764-3LX	300	300	120	5V $\pm$ 5%	C, I, L
AM2764-30DX AM2764-30LX	300	300	120	5V $\pm$ 10%	C, I, L
AM2764-4DX AM2764-4LX	450	450	150	5V $\pm$ 5%	C, I, L
AM2764-45DX AM2764-45LX	450	450	150	5V $\pm$ 10%	C, I, L, M
AM2764PC	250	250	100	5V $\pm$ 5%	C

Note: 1. X = C (0 to +70°C), X = I (-40 to +85°C), X = L (-55 to 100°C), X = M (-55 to +125°C).

# Advanced Micro Devices



## MOS Memories Am27128

### DISTINCTIVE CHARACTERISTICS

- Fast access time – 150ns, 200ns, 250ns, 300ns, 450ns
- Low power consumption
  - 525mW active
  - 130mW stand-by
- Single 5V power supply
- $\pm 10\%$   $V_{CC}$  supply tolerance available
- Fully static operation – no clocks
- Separate chip enable and output enable controls
- TTL compatible inputs/outputs
- 28-pin JEDEC approved 27128 pin-out
- Pin compatible to Am2764 EPROM and Am92128-128K ROM
- Fast programming time (3 min typical)

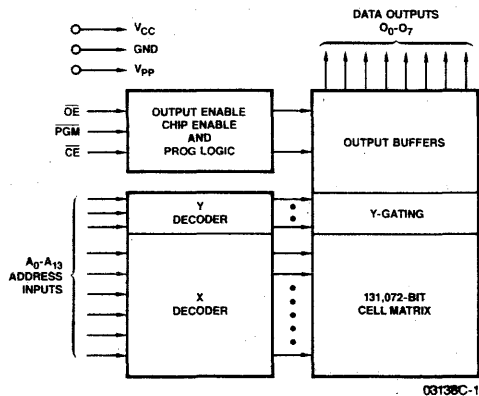
### GENERAL DESCRIPTION

The Am27128 is a 131,072-bit UV-light erasable and electrically programmable read-only memory. It is organized as 16384 words by 8-bits per word. The standard Am27128 offers access time of 250ns, allowing operation with high-speed microprocessors without any WAIT state.

To eliminate bus contention in a multiple-bus microprocessor system, the Am27128 offers separate output enable ( $\overline{OE}$ ) and chip enable ( $\overline{CE}$ ) controls.

All signals are TTL levels, including programming signals. Bit locations may be programmed singly, in blocks or at random. To reduce programming time, the Am27128 may be programmed using 1ms pulses. Typically, Am27128 can be programmed in three minutes. See Flow Chart on page 6 for details.

### BLOCK DIAGRAM



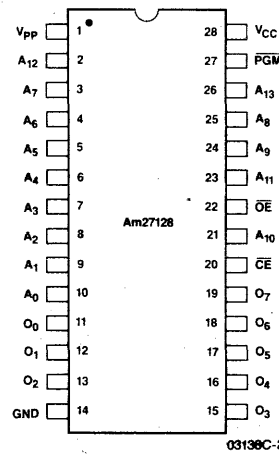
### MODE SELECTION

Mode	Pins	$\overline{CE}$ (20)	$\overline{OE}$ (22)	PGM (27)	$V_{pp}$ (1)	$V_{CC}$ (28)	Outputs (11-13, 15-19)
Read		$V_{IL}$	$V_{IL}$	$V_{IH}$	$V_{CC}$	$V_{CC}$	DOUT
Standby		$V_{IH}$	X	X	$V_{CC}$	$V_{CC}$	High Z
Program		$V_{IL}$	$V_{IH}$	$V_{IL}$	$V_{pp}$	$V_{CC}$	DIN
Program Verify		$V_{IL}$	$V_{IL}$	$V_{IH}$	$V_{pp}$	$V_{CC}$	DOUT
Program Inhibit		$V_{IH}$	X	X	$V_{pp}$	$V_{CC}$	High Z

X can be either  $V_{IL}$  or  $V_{IH}$

Figure 1.

### CONNECTION DIAGRAM – Top View



Note: Pin 1 is marked for orientation.

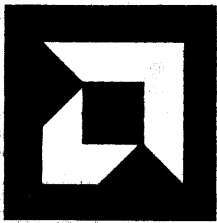
$A_0$ - $A_{13}$ : Addresses  
 $O_0$ - $O_7$ : Outputs  
 $\overline{CE}$ : Chip Enable  
 $\overline{OE}$ : Output Enable  
 PGM: Program

Figure 2.

### ORDERING INFORMATION

Order <sup>1</sup> Number	$t_{ACC}$ (ns)	$t_{CE}$ (ns)	$t_{OE}$ (ns)	$V_{CC}$	Temperature Range
AM27128-1DX	150	150	75	5V $\pm$ 5%	C, I
AM27128-15DX	150	150	75	5V $\pm$ 10%	C, I
AM27128-2DX	200	200	75	5V $\pm$ 5%	C, I, L
AM27128-20DX	200	200	75	5V $\pm$ 10%	C, I, L, M
AM27128DX	250	250	100	5V $\pm$ 5%	C, I, L
AM27128-25DX	250	250	100	5V $\pm$ 10%	C, I, L, M
AM27128-3DX	300	300	120	5V $\pm$ 5%	C, I, L
AM27128-30DX	300	300	120	5V $\pm$ 10%	C, I, L
AM27128-4DX	450	450	150	5V $\pm$ 5%	C, I, L
AM27128-45DX	450	450	150	5V $\pm$ 10%	C, I, L, M

Note: 1. X = C (0 to +70°C), X = I (-40 to +85°C), X = L (-55 to +100°C), X = M (-55 to +125°C).



# Advanced Micro Devices

## MOS Memories Am27256

### DISTINCTIVE CHARACTERISTICS

- Fast access time – 170ns, 200ns, 250ns, 300ns, 450ns
- Programming voltage: 12.5V
- Low power consumption
  - 525mW active
  - 130mW stand-by
- Single 5V power supply
- $\pm 10\%$   $V_{CC}$  supply tolerance available
- Fully static operation – no clocks
- Separate chip enable and output enable controls
- TTL compatible inputs/outputs
- 28-pin JEDEC approved Am27256 pin-out
- Pin compatible to Am2764 EPROM, Am27128 EPROM and Am92256 – 256K ROM
- Fast programming time (5 min typical)
- Auto select mode for automated programming

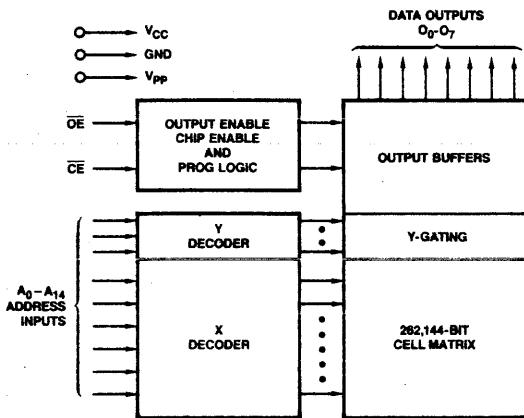
### GENERAL DESCRIPTION

The Am27256 is a 262,144 bit UV-light erasable and electrically programmable read-only memory. It is organized as 32,768 words by 8-bits per word. The standard Am27256 offers access time of 250ns, allowing operation with high-speed microprocessors without any WAIT state.

To eliminate bus contention on a multiple-bus microprocessor system, the Am27256 offers separate output enable ( $\overline{OE}$ ) and chip enable ( $\overline{CE}$ ) controls.

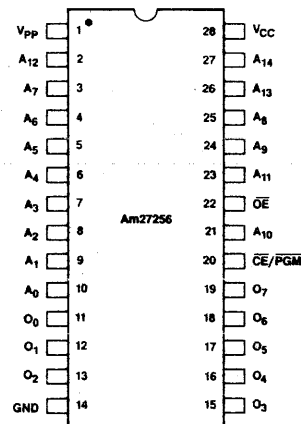
All signals are TTL levels, including programming signals. Bit locations may be programmed singly, in blocks or at random. To reduce programming time, the Am27256 may be programmed using 1ms pulses. Typically, the Am27256 can be programmed in five minutes.

### BLOCK DIAGRAM



040698-1

### CONNECTION DIAGRAM – Top View



Note: Pin 1 is marked for orientation. 040698-2

$A_0$ – $A_{14}$ : Addresses       $\overline{OE}$ : Output Enable  
 $O_0$ – $O_7$ : Outputs       $\overline{CE/PGM}$ : Chip Enable/Program

Figure 2.

Figure 1. Mode Select

Mode	Pins	$\overline{CE}$ (20)	$\overline{OE}$ (22)	$A_9$ (24)	$V_{PP}$ (1)	$V_{CC}$ (28)	Outputs (11-13, 15-19)
Read		$V_{IL}$	$V_{IL}$	X	$V_{CC}$	$V_{CC}$	DOUT
Output Disable		$V_{IL}$	$V_{IH}$	X	$V_{CC}$	$V_{CC}$	High Z
Stand By		$V_{IH}$	X	X	$V_{CC}$	$V_{CC}$	High Z
Program		$V_{IL}$	$V_{IH}$	X	$V_{PP}$	$V_{CC}$	DIN
Program Verify		$V_{IH}$	$V_{IL}$	X	$V_{PP}$	$V_{CC}$	DOUT
Program Inhibit		$V_{IH}$	$V_{IH}$	X	$V_{PP}$	$V_{CC}$	High Z
Auto Select		$V_{IL}$	$V_{IL}$	$V_H$	$V_{CC}$	$V_{CC}$	Code

Note: X can be either  $V_{IL}$  or  $V_{IH}$   
 $V_H = 12.0V \pm 0.5V$

### ORDERING INFORMATION

Order Number	$t_{ACC}$ (ns)	$t_{CE}$ (ns)	$t_{OE}$ (ns)	$V_{CC}$
Am27256-1DC	170	170	75	$5V \pm 5\%$
Am27256-2DC	200	200	75	$5V \pm 5\%$
Am27256-20DC	200	200	75	$5V \pm 10\%$
Am27256DC	250	250	100	$5V \pm 5\%$
Am27256-25DC	250	250	100	$5V \pm 10\%$
Am27256-3DC	300	300	120	$5V \pm 5\%$
Am27256-30DC	300	300	120	$5V \pm 10\%$
Am27256-4DC	450	450	150	$5V \pm 5\%$
Am27256-45DC	450	450	150	$5V \pm 10\%$

# ABBREVIATIONS OF COMPANY NAMES

<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.	<b>Harris</b>	Harris Semiconductor	<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Heurikon</b>	Heurikon Corp.	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hitachi</b>	Hitachi America, Ltd.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Holt</b>	Holt Inc.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>HP</b>	Hewlett-Packard		
<b>Analogic</b>	Analogic	<b>Hughes</b>	Hughes Aircraft, Solid State Products	<b>Panasonic</b>	Panasonic
<b>Analog Sys</b>	Analog Systems			<b>Pico Design</b>	Pico Design
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Polycore</b>	Polycore Electronics
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>Plessey</b>	Plessey Semiconductors
<b>APM</b>	Applied Microsystems Corp.			<b>PMI</b>	Precision Monolithics, Inc.
<b>Appl Sys</b>	Applied Systems Corp.	<b>ICC</b>	International Cybernetics	<b>PragDes</b>	Pragmatic Design Inc.
<b>APT</b>	Applied Microtechnology	<b>IDT</b>	Integrated Device Technology	<b>Pro-Log</b>	Pro-Log Corp.
<b>Aptek</b>	Aptek Microsystems	<b>IMI</b>	International Microcircuits, Inc.		
<b>Array Tech</b>	Array Technology	<b>IMP</b>	International Microelectronic Products		
<b>AWI</b>	AWI Electronics		Industrial MicroSystems Inc.	<b>Raytheon</b>	Raytheon Semiconductor
		<b>IMS</b>	Infosphere	<b>RCA</b>	RCA Solid State Division
<b>Barvon</b>	Barvon Research	<b>Infosphere</b>	Infosphere	<b>RCI Data</b>	RCI Data
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>Inmos</b>	Inmos	<b>RELMS</b>	Relational Memory Systems
<b>Burr-Brown</b>	Burr-Brown	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>Reticon</b>	Reticon
		<b>IntCirSys</b>	Integrated Circuit Systems	<b>RIFA</b>	RIFA
<b>CAE</b>	Computer Aided Engineering	<b>IntCompSys</b>	Integrated Computer Systems	<b>Rockwell</b>	Rockwell, Microelectronic Devices
<b>Cal Devices</b>	California Devices	<b>Int Tech</b>	Integrated Technology Corp.	<b>RTC</b>	Riehl Time Corporation
<b>Cermetek</b>	Cermetek	<b>Intech</b>	Intech Microcircuits Div.		
<b>CGRS</b>	CGRS Microtech Inc.	<b>Intel</b>	Intel		
<b>Cherry</b>	Cherry Semiconductor	<b>Interdesign</b>	Interdesign	<b>Sanyo</b>	Sanyo
<b>CIC</b>	Custom Integrated Circuits	<b>Intersil</b>	Intersil	<b>SBE</b>	SBE, Inc.
<b>CirTech</b>	Circuit Technology	<b>Intronics</b>	Intronics	<b>SEEQ</b>	SEEQ Technology, Inc.
<b>Citel</b>	Citel	<b>ITT</b>	ITT Semiconductors	<b>SPI</b>	Semi Processes Inc.
<b>Comlinear</b>	Comlinear Corporation			<b>Siemens</b>	Siemens
<b>CMA</b>	Custom MOS Arrays	<b>Kinetic Sys</b>	Kinetic Systems	<b>Si-Fab</b>	Si-Fab
<b>Comark</b>	Comark Corp.	<b>Kontron</b>	Kontron Electronics	<b>Signetics</b>	Signetics
<b>Comdial</b>	Comdial Semiconductor			<b>SGS</b>	SGS Semiconductor
<b>Comp Auto</b>	Computer Automation	<b>Lambda</b>	Lambda Semiconductor	<b>Sharp</b>	Sharp
<b>Compas</b>	Compas Microsystems	<b>Linear Tech</b>	Linear Technology	<b>Silicon G</b>	Silicon General
<b>Cont Logic</b>	Control Logic Inc.	<b>LSI Comp</b>	LSI Computer Systems	<b>Siliconix</b>	Siliconix
<b>Control Sys</b>	Control Systems Microsystems Div.	<b>LSI Logic</b>	LSI Logic Corporation	<b>Silicon Sys</b>	Silicon Systems Inc.
<b>CreMicro</b>	Creative Micro Systems			<b>Siltronics</b>	Siltronics
<b>Cromemco</b>	Cromemco, Inc.	<b>Master Logic</b>	Master Logic Corporation	<b>SMC</b>	Standard Microsystems Corp.
<b>Cubit</b>	Cubit Inc.	<b>Matrix</b>	Matrix Corp.	<b>S MOS</b>	S MOS Systems
<b>Curtis</b>	Curtis Electro Devices, Inc.	<b>Matrox</b>	Matrox Electronic Systems	<b>Solarise</b>	Solarise Enterprises
<b>Cybernetic</b>	Cybernetic Micro Systems	<b>Maxim</b>	Maxim Integrated Products	<b>Solitron</b>	Solitron Devices
<b>Cybersys</b>	Cybersystems	<b>MCC</b>	Micro-Computer Control	<b>Sprague</b>	Sprague Electric Company
<b>Cybertek</b>	Cybertek Inc.	<b>MCE</b>	MCE Electronics	<b>SSM</b>	Solid State Micro Technology
		<b>Micrel</b>	Micrel		for Music
<b>Data General</b>	Data General	<b>Micro Innov</b>	Micro Innovators	<b>SSS</b>	Solid State Scientific
<b>Data I/O</b>	Data I/O	<b>Micropac</b>	Micropac Industries	<b>Stag</b>	Stag Microsystems
<b>Data Trans</b>	Data Translation	<b>Micro Net</b>	Micro Networks	<b>STC</b>	Storage Technology Corp.
<b>Datel</b>	Datel	<b>Micro Pwr</b>	Micro Power Systems	<b>STD</b>	STD Microsystems
<b>Datricon</b>	Datricon Corporation	<b>Micro Sci</b>	Micro Sciences Corp.	<b>Struc Des</b>	Structured Design Inc.
<b>DDC</b>	Data Devices Corporation	<b>Micro Tech</b>	Microcircuits Technology	<b>Stynetic</b>	Stynetic Systems
<b>DEC</b>	Digital Equipment Corporation	<b>Micro-Link</b>	Micro-Link Corporation	<b>Sunrise</b>	Sunrise Electronics
<b>Die-Tech</b>	Die-Tech	<b>Micron</b>	Micron Technology	<b>Sunshine</b>	Sunshine Semiconductor
<b>Digelec</b>	Digelec Corp.	<b>MillerTron</b>	MillerTronics	<b>Supertex</b>	Supertex Inc.
<b>Digitek</b>	Digitek, Inc.	<b>Miller</b>	Miller Technology	<b>Symtek</b>	Symtek Corp.
<b>Dionics</b>	Dionics Inc.	<b>Mitel</b>	Mitel Semiconductor	<b>Synertek</b>	Synertek
<b>Dist Comp</b>	Distributed Computer Systems	<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Sys Innov</b>	Systems Innovations
<b>Divers Tech</b>	Diversified Technology	<b>MMI</b>	Monolithic Memories, Inc.		
		<b>Mostek</b>	Monolithic Systems Corp.	<b>Tau Zero</b>	Tau Zero Inc.
<b>E-HI</b>	E-H International, Inc.	<b>Motorola</b>	Motorola Semiconductor	<b>Technitrol</b>	Technitrol
<b>EDI</b>	Electronic Designs Inc.	<b>MRC</b>	MRC Systems	<b>Tektronix</b>	Tektronix
<b>Elind</b>	Elind Electronica Industriale	<b>Murray</b>	Murray Consulting	<b>Teledyne C</b>	Teledyne Crystalonics
<b>EEM-SESCO</b>	EEM-SESCO	<b>Monesil</b>	Monesil	<b>Teledyne P</b>	Teledyne Philbrick
<b>Emulogic</b>	Emulogic Inc.			<b>Teledyne S</b>	Teledyne Semiconductor
<b>ETI Micro</b>	ETI Micro	<b>National</b>	National Semiconductor	<b>Telefunken</b>	Telefunken
<b>Exar</b>	Exar Integrated Systems	<b>NCM</b>	NCM Corp.	<b>Telmos</b>	Telmos
<b>Exel</b>	Exel Microelectronics	<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Teltone</b>	Telton Corporation
		<b>NEC</b>	NEC Electronics	<b>TI</b>	Texas Instruments
<b>Fairchild</b>	Fairchild	<b>Nitron</b>	Nitron	<b>Third Domain</b>	Third Domain
<b>Ferranti</b>	Ferranti Electric			<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
<b>Force</b>	Force Computers			<b>Toshiba</b>	Toshiba America
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<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.			<b>TRW</b>	TRW LSI Products
				<b>Unitrode</b>	Unitrode
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				<b>Xicor</b>	Xicor, Inc.
				<b>Xycom</b>	Xycom
				<b>Zendex</b>	Zendex Corp.
				<b>Zilog</b>	Zilog
				<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrex</b>	Zytrex Corp.



## 32,768 BIT (4096x8) STATIC NMOS ROM

### Features

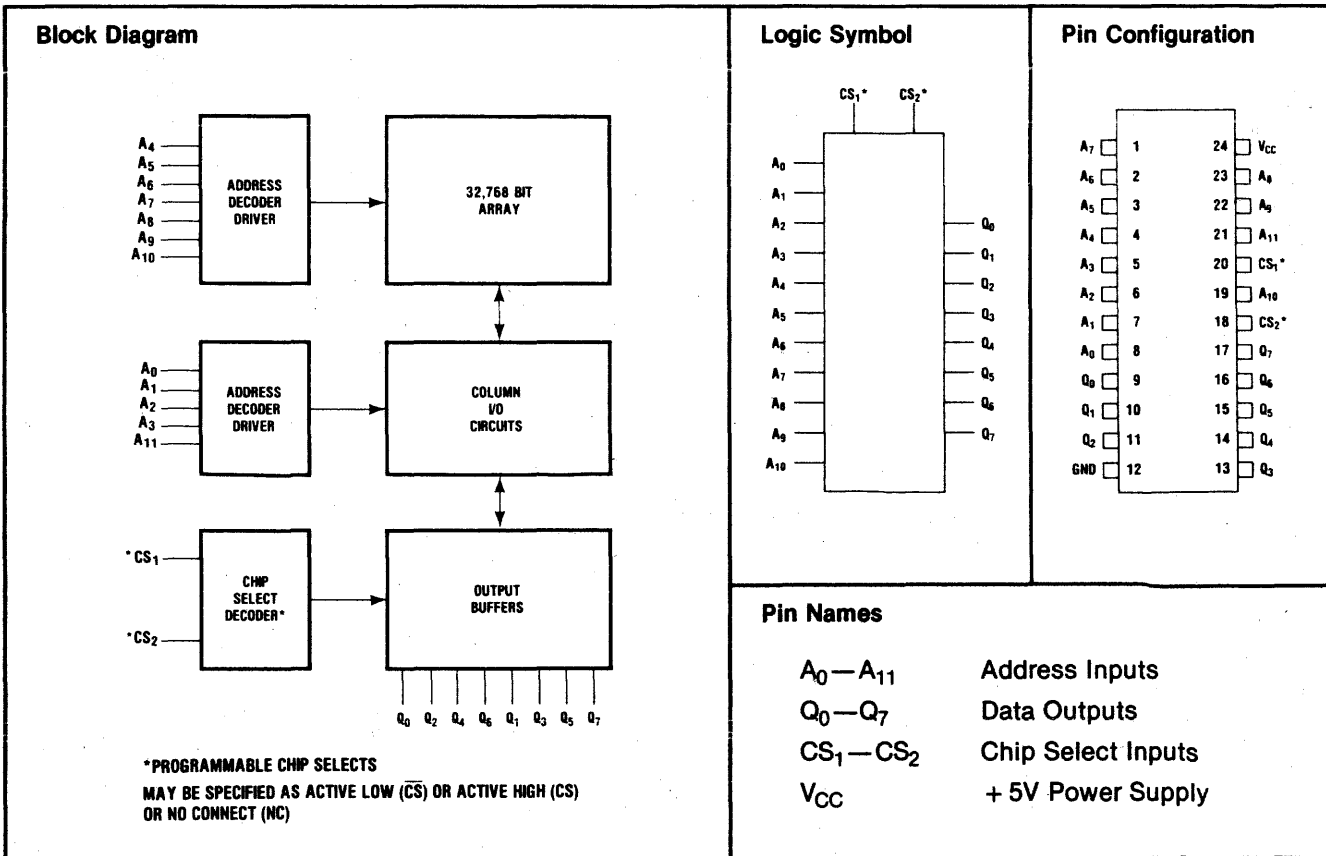
- Fast Access Time: 350ns Maximum
- Fully Static Operation
- Single +5V ±5% Power Supply
- Directly TTL Compatible Inputs
- Three-State TTL Compatible Outputs
- Two Programmable Chip Selects
- EPROM Pin Compatible (2732)
- Extended Temperature Range Available

### General Description

The AMI S2333 is a 32,768 bit static mask programmable NMOS ROM organized as 4096 words by 8 bits. The device is fully TTL compatible on all inputs and outputs and has a single +5V power supply. The three state outputs facilitate memory expansion by allowing the outputs to be OR-tied to other devices.

The S2333 is pin compatible with UV EPROMs making system development much easier and more cost effective. The fully static S2333 requires no clocks for operation. The two chip selects are mask programmable with the active level for each being specified by the user.

The S2333 is fabricated using AMI's N-Channel MOS technology. This permits the manufacture of very high density, high performance mask programmable ROMs.





## 65,536 BIT (8192x8) STATIC NMOS ROM

### Features

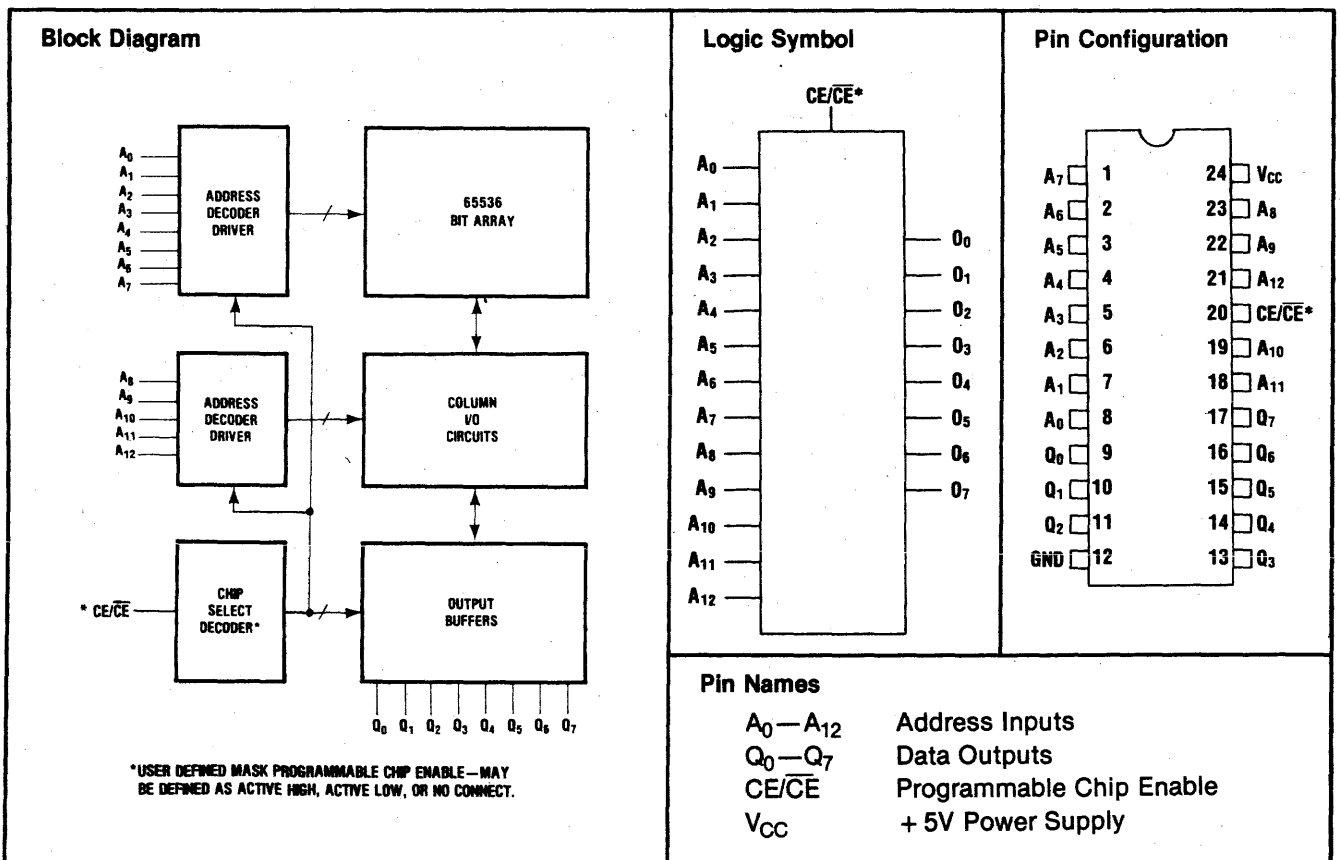
- Fast Access Time: S68A364-350ns Maximum  
S58B364-250ns Maximum
- Low Standby Power: 85mW Maximum
- Late Mask Programmable
- Fully Static Operation
- Single +5V  $\pm$ 10% Power Supply
- Directly TTL Compatible Inputs
- Three-State TTL Compatible Outputs
- Programmable Chip Enable

### General Description

The AMI S68364 family are 65,536 bit static mask programmable NMOS ROMs organized as 8192 words by 8 bits. The devices are fully TTL compatible on all inputs and outputs and have a single +5V power supply. The three-state outputs facilitate memory expansion by allowing the outputs to be OR-tied to other devices.

The devices are fully static, requiring no clocks for operation. The chip enable is mask programmable, the active level being specified by the user. When not enabled, power supply current is reduced to a maximum of 15mA.

The S68364 family of devices are fabricated using AMI's NMOS ROM technology. This permits the mask programmable ROMs.





## S2364A/S2364B

# 65,536 BIT (8192x8) STATIC NMOS ROM

### Features

- Fast Access Time: S2364A 350ns Maximum  
S2364B 250ns Maximum
- Low Standby Power: 85mW Maximum
- Fully Static Operation
- Single +5V ± 10% Power Supply
- Directly TTL Compatible Inputs
- Three-State TTL Compatible Outputs
- Three Programmable Chip Enables/Selects
- EPROM Pin Compatible (2764)
- Late Mask Programmable

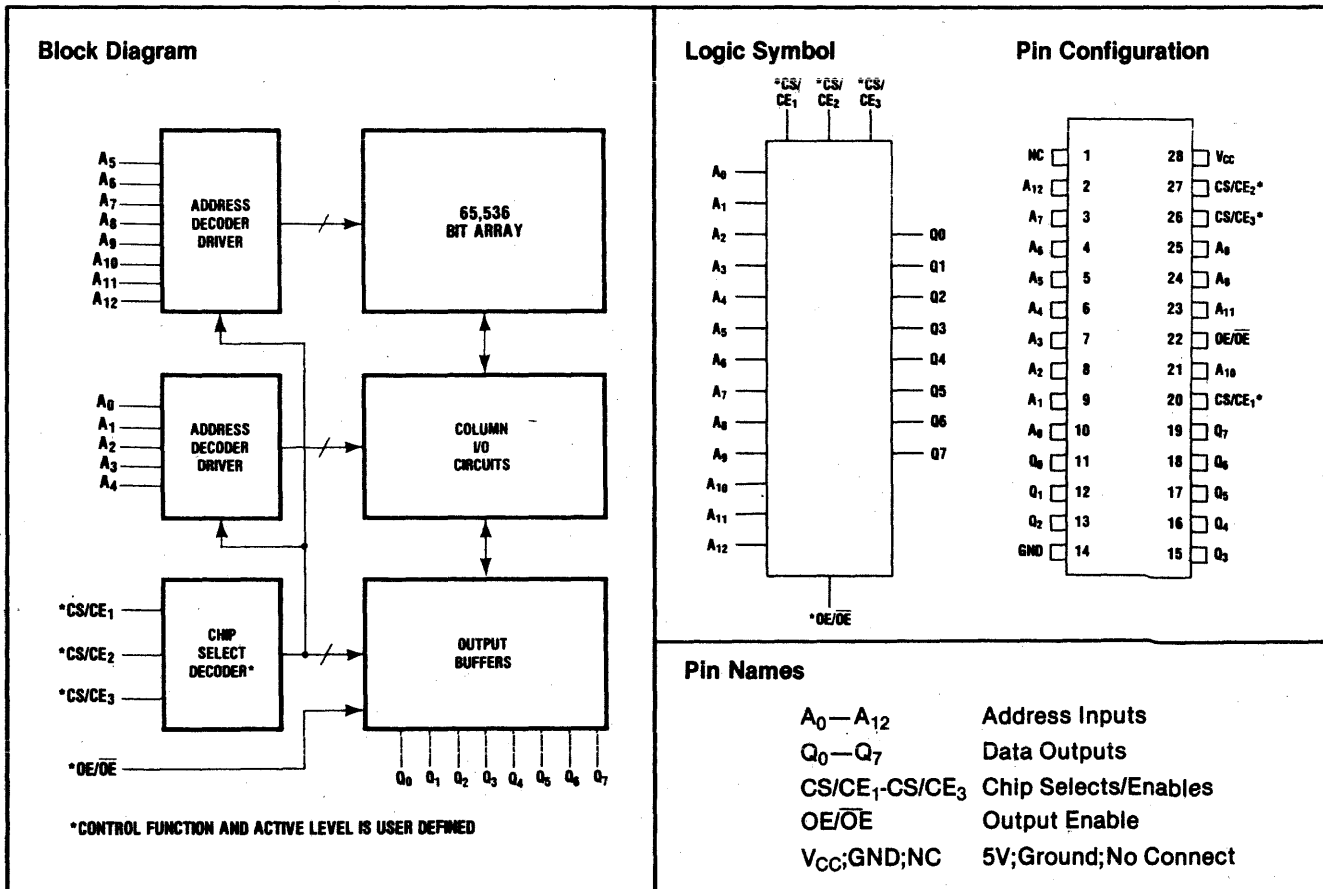
### General Description

The AMI S2364 is a 65,536 bit static mask programmable NMOS ROM organized as 8192 words by 8 bits.

The device is fully TTL compatible on all inputs and outputs and has a single +5V power supply. The three state outputs facilitate memory expansion by allowing the outputs to be OR-tied to other devices.

The S2364 is pin compatible with the 2764 UV EPROM making system development much easier and more cost effective. It is fully static, requiring no clocks for operation. The three chip enables are mask programmable; the active level for each being specified by the user. When not enabled, power supply current is reduced to a 10mA maximum.

The S2364 is fabricated using AMI's N-Channel MOS technology. This permits the manufacture of very high density, high performance mask programmable ROMs.





# 131,072 BIT (16384x8) STATIC NMOS ROM

### Features

- Fast Access Time: S23128A-350ns Maximum  
S23128B-250ns Maximum
- Low Standby Power: 110mW Max.
- Fully Static Operation
- Single +5V ± 10% Power Supply
- Directly TTL Compatible Outputs
- Three-State TTL Compatible Outputs
- Two Programmable Chip Enables/Selects
- EPROM Pin Compatible (27128)
- Late Mask Programmable
- Programmable Output/Chip Enable

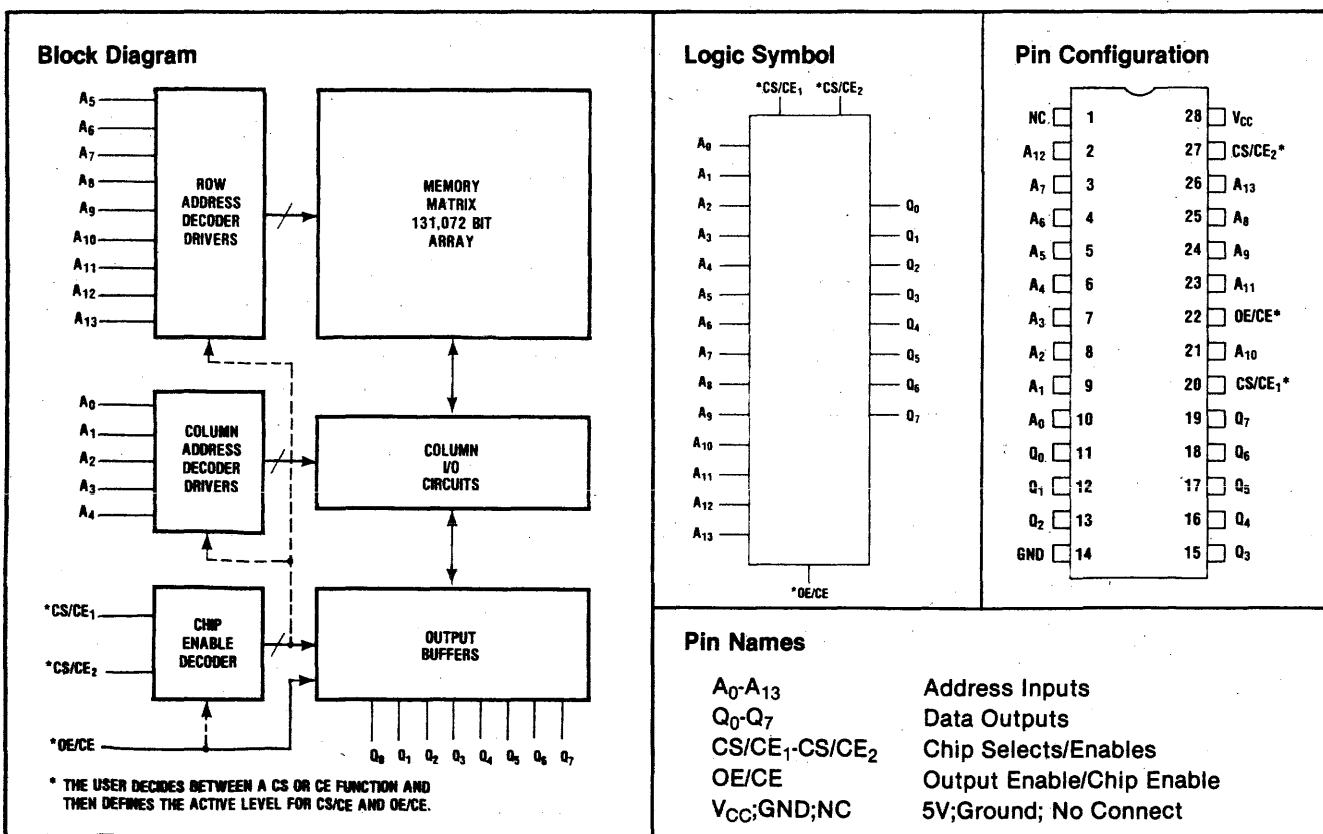
### General Description

The AMI S23128 is a 131,072 bit static mask programmable NMOS ROM organized as 16,384 words by 8 bits.

The device is fully TTL compatible on all inputs and outputs and has a single +5V power supply. The three state outputs facilitate memory expansion by allowing the outputs to be OR-tied to other devices.

The S23128 is pin compatible with the 27128 EPROM making system development easier and more cost effective. The fully static S23128 requires no clocks for operation. The three control pins are mask programmable with the active level and function being specified by the user. The pins can also be programmed as no connections. If CE functions are selected, automatic powerdown is available. The power supply current is reduced to 12mA when the chip is disabled.

The S23128 is fabricated using AMI's NMOS technology. This permits the manufacture of high density, high performance ROMs.





## 262,144 BIT (32,768x8) STATIC NMOS ROM

### Features

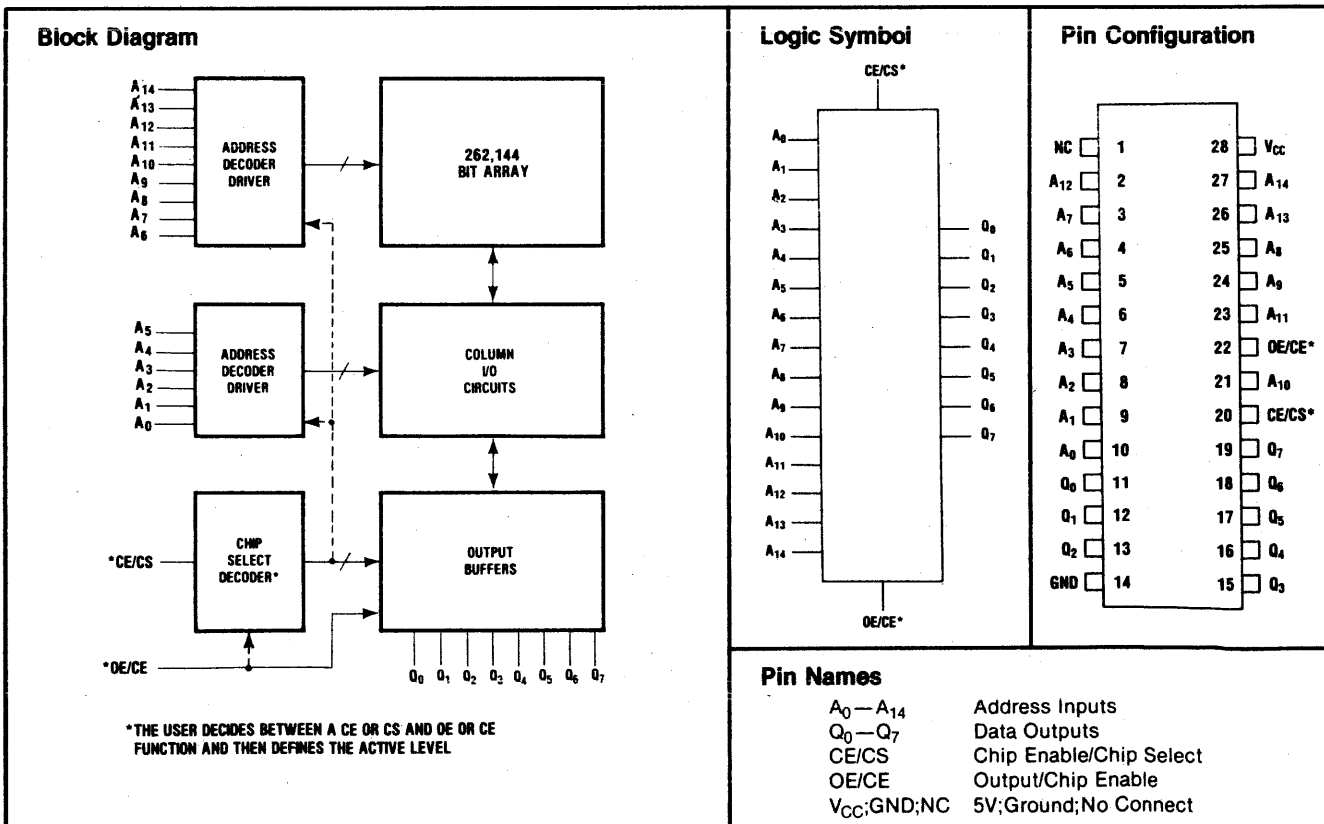
- Fast Access Time:
  - S23256B: 250ns Maximum
  - S23256C: 150ns Maximum
- Low Power Dissipation
  - Active Current: 40mA Maximum
  - Standby Current: 10mA Maximum
- Fully Static Operation
- Two User-Defined and Programmable Control Lines: CE/CS, OE/CE
- EPROM Pin Compatible
- Late Mask Programmable
- Three-State TTL Compatible Outputs

### General Description

The AMI S23256 is a 262,144 bit static mask programmable NMOS ROM organized as 32,768 words by 8 bits. The devices are fully TTL compatible on all inputs and outputs and operate from a single +5V ± 10% power supply. The three state outputs facilitate memory expansion by allowing the outputs to be OR-tied to other devices.

The S23256 is pin compatible with the 27128 UV EPROM making system development much easier and more cost effective. It is fully static, requiring no clocks for operation.

The S23256 is fabricated using AMI's N-Channel MOS ROM technology. This permits the manufacture of very high density, high performance mask programmable ROMs.



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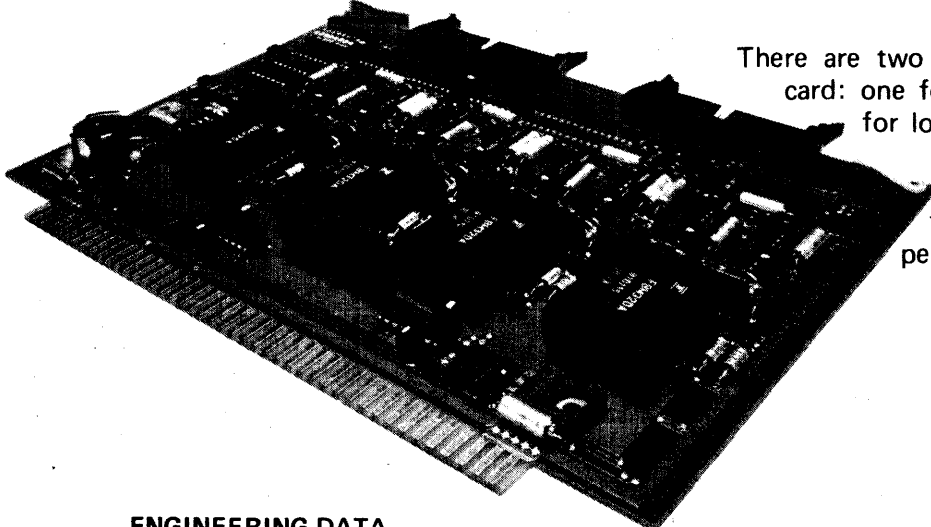
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# 32K byte Bubble Memory Card



There are two kinds of Fujitsu bubble memory card: one for high-speed file memory and one for low-speed file memory.

Both can be connected to the 8-bit micro-processor buss and the capacity is easily increased depending on a system.

## ENGINEERING DATA

	FBC304M1A	FBC304D2A
Devices	FBM31DB	FBM32DA
Number of Devices	4	4
Organization	Serial loop	Major/Minor loop
Capacity	296,128 bits	324,024 bits
Average Access Time	370 ms	4.5 ms
Drive Frequency	100KHz	100KHz
Data Rate	100K bits/sec.	200K bits/sec.
Interface	TTL Compatible	TTL Compatible
Power Consumption		
Operating	9 W	11 W
Stand-by	6 W	2.5 W
Power Requirement *	+ 12 V (± 5 %) 0.3A Max. -12 V (± 5 %) 0.2A Max. + 5 V (± 5 %) 1.6A Max.	+ 12 V (± 5 %) 1.0A Max. -12 V (± 5 %) 0.2A Max. + 5 V (± 5 %) 0.7A Max.
Temperature Range		
Operating	0°C to +50°C	0°C to +50°C
Non-volatile Storage	-40°C to +85°C	-40°C to +85°C
Card Dimension	230 x 160 mm <sup>2</sup>	230 x 160 mm <sup>2</sup>

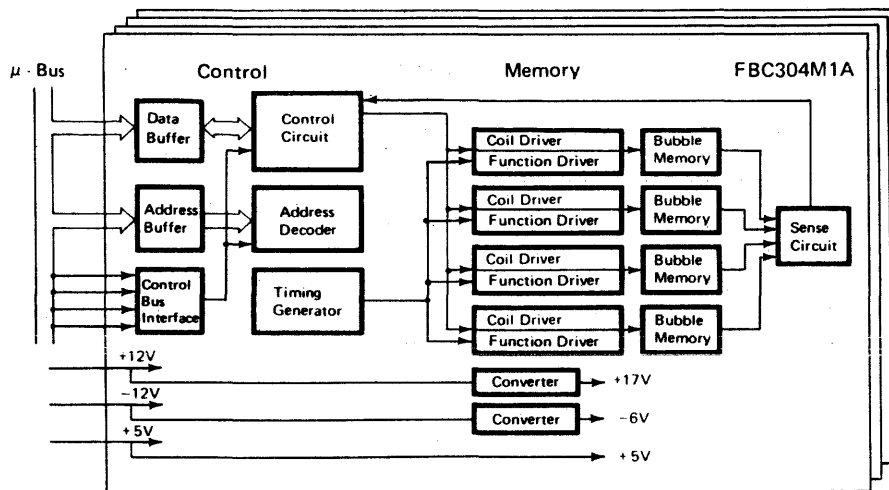
\* Sequential power-on and power-off are unnecessary.

## Fujitsu 32K byte Bubble Memory Card

FBC304M1A is a 32K byte bubble memory card containing four 74K bit serial loop organized Bubble Memory Devices FBM31DB.

The timing and control circuits, in addition to

coil drivers, function drivers, sense amplifier, etc., are also mounted on the same card, which can be directly connected to the 8-bit micro-processor as a low-speed data file and program memory.

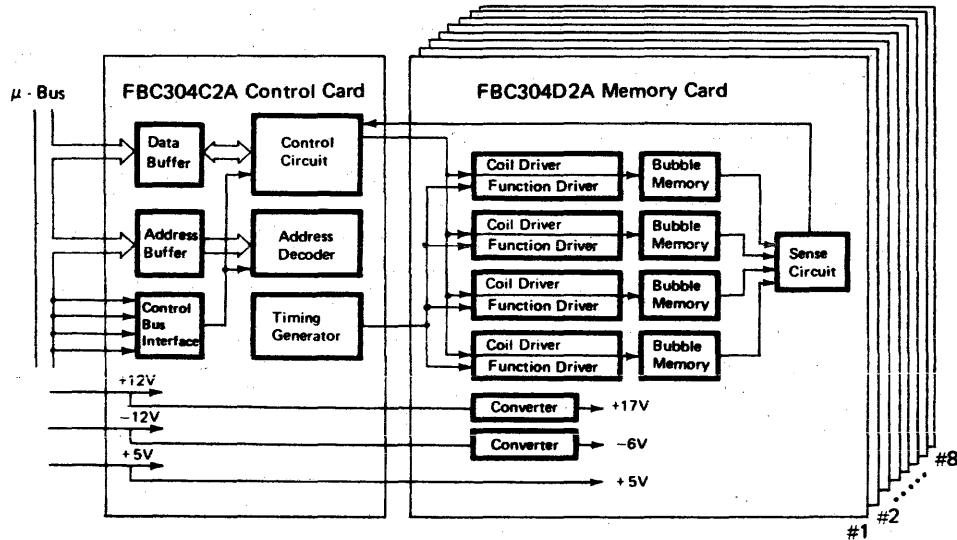


Fujitsu America, Inc., 918 Sherwood Drive, Lake Bluff, Illinois 60044

Fujitsu America  
MEMORY

FBC304D2A is a 32K byte high-speed bubble memory card with a TTL interface, containing four 83K bit major/minor loop organized Bubble Memory Devices FBM32DA with peripheral circuits. As a high-speed random access file memory

this card can be connected to the 8 bit micro-processor buss via a control card FBC304C2A and the number of cards can be increased up to 256K bytes (8 cards), according to a system scale.



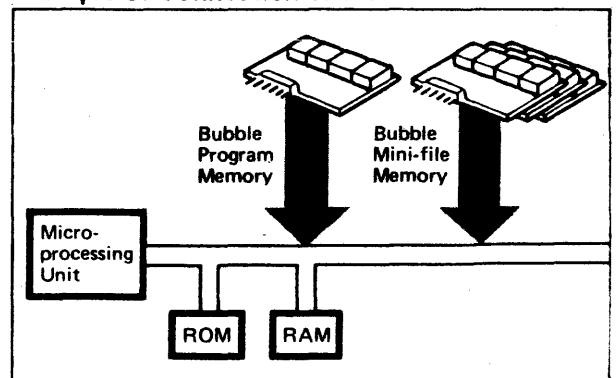
**EXPECTED BUBBLE MEMORY APPLICATIONS**

System Memory Capacity (Byte)	10K	100K	1M	10M
Chip Capacity (bit)	64K ~ 100K	256K ~ 1M	> 1 M	
Bubble Memory Products	Program Memory 	Mini-File Memory 	File Memory 	
Competing Technology (Application Area)	EP ROM 	Floppy Disc MT Cassette 	Drum Mini-disc 	

**Applications**

- File memory for micro-computer and mini-computer
- Replacement for disc, drum and tape devices
- Program loaders for testing equipment and numerical control systems

**Example of Connection to a CPU**



# 64K bit Bubble Memory Device



FBM31DB and FBM32DA are 18-pin DIP packages that contain a 64K bit serial loop chip and a 64K bit major/minor loop chip, respectively, orthogonal coils for rotating magnetic field, and a magnet for biasing which makes the bubbles stable to keep

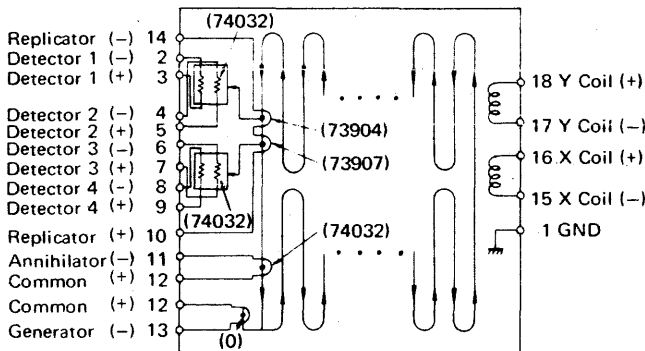
data. These bubble memory devices serve as the heart of the Fujitsu bubble memory card mounted on a printed circuit board.

## ENGINEERING DATA

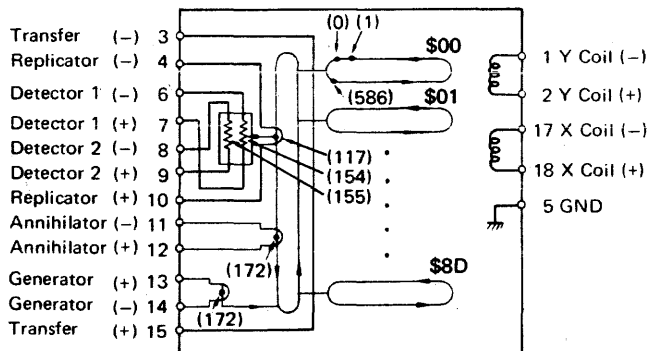
	FBM31DB	FBM32DA
Capacity:	74,032 bits	83,354 bits (total) 81,006 bits (effective)
Organization:	Serial loop	Major/Minor loop
Drive Frequency:	100KHz	100KHz
Transfer Rate:	100K bits/sec.	50K bits/sec.
Average Access Time:	370 ms	4.5 ms
Power Consumption:	500 mW	500 mW
Temperature Range		
Operating:	0°C to +55°C	0°C to +55°C
Non-volatile Storage:	-40°C to +85°C	-40°C to +85°C
External Magnetic Field		
Operating:	50 Oe Max.	50 Oe Max.
Non-Volatile Storage:	100 Oe Max.	100 Oe Max.
Physical Structure:	18 pin DIP	18 pin DIP
Dimensions:	27.94 x 31 x 10 mm <sup>3</sup>	27.94 x 31 x 10 mm <sup>3</sup>
Weight	30 g	30 g

## CIRCUIT DIAGRAM

### ● FBM31DB (Serial loop)



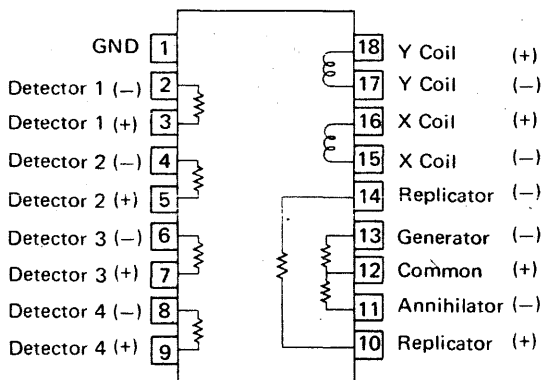
### ● FBM32DA (Major/Minor loop)



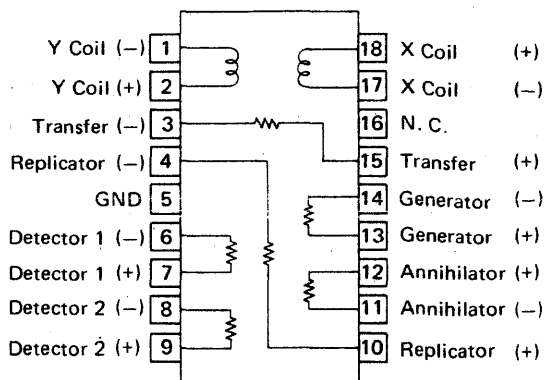
\* Loop numbers (00 through 8D) are designated by hexadecimal code.

## PIN ASSIGNMENT

### ● FBM31DB (Serial loop)



### ● FBM32DA (Major/Minor loop)



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# 256K bit Bubble Memory Device

FUJITSU

The FBM42DA is a major/minor bubble memory device in a 16 DIP package containing a 256K bit chip; the FBM43DA is a block replicator transfer bubble memory device in a 20 pin DIP package containing a 256K bit chip. Both devices are

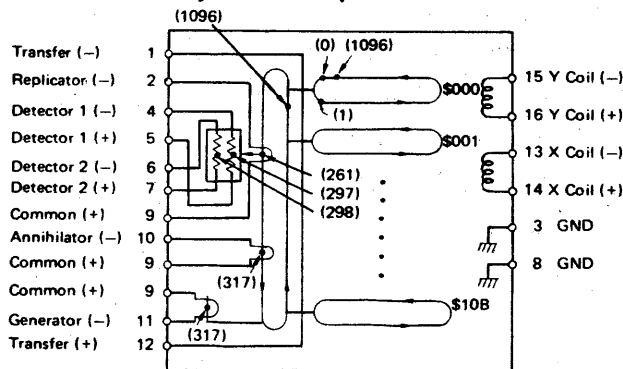
mountable on a printed circuit board and have orthogonal coils for rotating magnetic field. They also incorporate magnets for biasing, thereby providing the bubbles with stability to retain data.

## ENGINEERING DATA

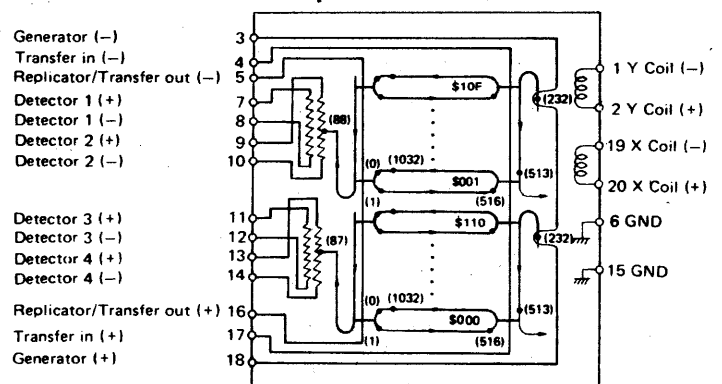
	FBM42DA	FBM43DA
Capacity	283,026 bits (effective) 1,097 bits X 258 loops	273,745 bits (effective) 1,033 bits X 265 loops
Organization	Major/minor loop	Major/minor loop (Block replicator transfer system)
Drive Frequency	100KHz	100KHz
Transfer Rate	50K bits/sec	100K bits/sec
Average Access Time	8.5 ms	6.0 ms
Power Consumption	670 mW	670 mW
Temperature Range Operating Non-volatile Storage	0°C to +55°C -40°C to +85°C	0°C to +55°C -40°C to +85°C
External Magnetic field	50 Oe Max.	50 Oe Max.
Physical Structure	16 pin DIP	20 pin DIP
Dimensions	27.94 x 31 x 11 mm <sup>3</sup>	27.94 x 31 x 11 mm <sup>3</sup>
Weight	30 g	30 g

## CIRCUIT DIAGRAM

### ● FBM42DA (Major/minor loop)

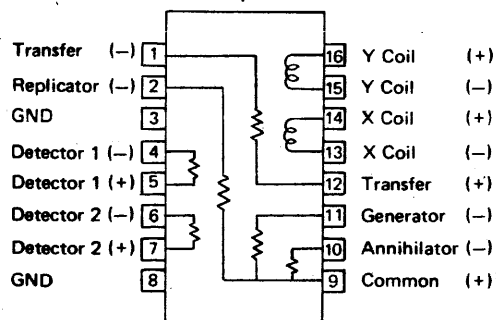


### ● FBM43DA (Block replicator transfer)

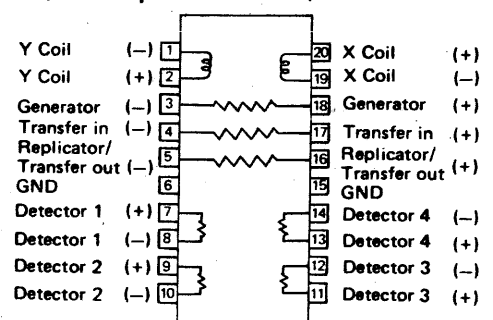


## PIN ASSIGNMENT

### ● FBM42DA (Major/minor loop)



### ● FBM43DA (Block replicator transfer)



\* The \$ mark prefixed hexadecimal codes in the circuit diagrams indicate minor loop numbers.

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### Description

The HM-7602/03 is a fully decoded high speed Schottky TTL 256/Bit Field Programmable ROM in a 32 word by 8 bit/word format with open collector (HM-7602) or "Three State" (HM-7603) outputs. These PROMs are available in a 16 pin D.I.P. (ceramic or power plastic).

All bits are manufactured storing a logical "1" (Positive Logic) and can be selectively programmed for a logical "0" in any one bit position.

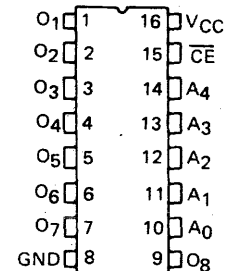
Nickel-chromium fuse technology is used on this and all other Harris Bipolar PROMs.

The HM-7602/03 contains test rows which are in addition to the storage array to assure high programmability and guarantee parametric and A.C. performance. The fuses in these test rows are blown prior to shipment.

There is one chip enable input on the HM-7602/03.  $\overline{CE}$  low enables the chip.

### Pinout

TOP VIEW — DIP



PIN NAMES

A<sub>0</sub> — A<sub>4</sub> Address Inputs  
 O<sub>1</sub> — O<sub>8</sub> Data Outputs  
 $\overline{CE}$  Chip Enable Input

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-7602/03-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )

HM-7602/03-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )

Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub> I <sub>IL</sub>	Address/Enable "1" Input Current "0"	—	—	+40 -250	$\mu A$ $\mu A$	$V_{IH} = V_{CC} \text{ Max.}$ $V_{IL} = 0.45V$
V <sub>IH</sub> V <sub>IL</sub>	Input Threshold "1" Voltage "0"	2.0 —	1.5 1.5	— 0.8	V V	$V_{CC} = V_{CC} \text{ Min.}$ $V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub> V <sub>OL</sub>	Output Voltage "1" "0"	2.4*	3.2*	— 0.45	V V	$I_{OH} = -2.0mA$ , $V_{CC} = V_{CC} \text{ Min.}$ $I_{OL} = +16mA$ , $V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub> I <sub>OLE</sub>	Output Disable "1" Current "0"	—	—	+100 -100	$\mu A$ $\mu A$	$V_{OH}$ , $V_{CC} = V_{CC} \text{ Max.}$ $V_{OL} = 0.3V$ , $V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	—	—	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15*	—	-100*	mA	$V_{CC} = V_{CC} \text{ Max.}$ , $V_{OUT} = 0.0V$ One Output Only for a Max. of 1 Second.
I <sub>CC</sub>	Power Supply Current	—	90	130	mA	$V_{CC} = V_{CC} \text{ Max.}$ All Inputs Grounded

NOTE: Positive current defined as into device terminals

\* "Three State" only

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical measurements are at  $T_A = +25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7602/03-5 5V $\pm 5\%$ 0°C to +75°C		HM-7602/03-2/-8 5V $\pm 10\%$ -55°C to +125°C		UNITS
		TYPICAL	MAXIMUM*	TYPICAL	MAXIMUM*	
T <sub>AA</sub>	Address Access Time	30	50	—	60	ns
T <sub>EA</sub>	Chip Enable Access Time	20	35	—	50	ns

\*A.C. limits guaranteed for worst case N<sup>2</sup> sequencing with maximum test frequency of 5MHz.

#### Description

The HM-7610/11 are fully decoded high speed Schottky TTL 1024 bit Field Programmable ROM's in a 256 word by 4 bit/word format with open collector (HM-7610) or "three state" (HM-7611) outputs. The PROMs are available in 16 pin D.I.P. (ceramic or power plastic).

All bits are manufactured storing a logical "1" (positive logic) and can be selectively programmed for a logical "0" in any bit position.

The HM-7610/11 contain test rows and columns which are in addition to the storage array to assure high programmability and guarantee parametric and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

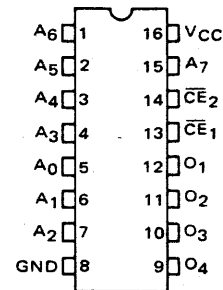
This PROM is intended for use in state of the art high speed logic systems.

Nickel-chromium fuse technology is used on these and all other Harris Bipolar PROMs.

There are two chip enable inputs on the HM-7610/11 where  $\overline{CE}_1$  and  $\overline{CE}_2$  low enables the chip.

#### Pinouts

TOP VIEW-DIP



#### PIN NAMES

- A<sub>0</sub> - A<sub>7</sub> Address Inputs
- O<sub>1</sub> - O<sub>4</sub> Data Outputs
- $\overline{CE}_1, \overline{CE}_2$  Chip Enable Inputs

**D.C. ELECTRICAL CHARACTERISTICS (Operating)** HM-7610/11-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-7610/11-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	1.5	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	1.5	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4*	3.2*	-	V	$I_{OH} = -2.0mA, V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA, V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}, V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40*	$\mu A$	$V_{OL} = 0.3V, V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current *	-15*	-	-100*	mA	$V_{CC} = V_{CC} \text{ Max.}, V_{OUT} = 0.0V$ One Output Only for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	90	110	mA	$V_{CC} = V_{CC} \text{ Max.}$ All Inputs Grounded

NOTE: Positive current defined as into device terminals.  
 \* "Three-State" only.

**A.C. ELECTRICAL CHARACTERISTICS (Operating)**  
 Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7610/11-5 5V $\pm 5\%$ 0°C to +75°C						HM-7610/11-2/-8 5V $\pm 10\%$ -55°C to +125°C						UNITS
		"B"		"A"		STD		"B"		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	25	35	35	45	40	60	-	50	-	65	-	75	ns
T <sub>EA</sub>	Chip Enable Access Time	-	25	-	25	-	25	-	30	-	30	-	30	ns
T <sub>DA</sub>	Chip Disable Access Time	-	25	-	25	-	25	-	30	-	30	-	30	

A.C. limits guaranteed for worst case N<sub>2</sub> sequencing with maximum test frequency of 5MHz.

### Description

The HM-7620/21 are fully decoded high speed Schottky TTL 2048 bit Field Programmable ROM's in a 512 word by 4 bit/word format with open collector (HM-7620) or "three state" (HM-7621) outputs. These PROMs are available in 16 pin D.I.P. (ceramic or power plastic).

All bits are manufactured storing a logical "1" (positive logic) and can be selectively programmed for a logical "0" in any bit position.

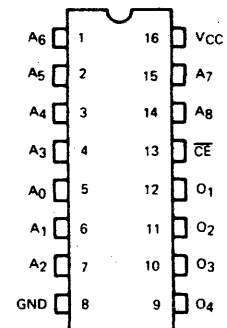
The HM-7620/21 contain test rows and columns which are in addition to the storage array to assure high programmability and guarantee parametric and A. C. performance. The fuses in these test rows and columns are blown prior to shipment.

This PROM is intended for use in state of the art high speed logic systems. Nickel-chromium fuse technology is used on these and all other Harris Bipolar PROMs.

There is a single chip enable input on the HM-7620/21 where  $\overline{CE}$  low enables the chip.

### Pinout

TOP VIEW - DIP



PIN NAMES

A0 - A8    Address Inputs  
 $\overline{CE}$     Chip Enable Input  
 O1 - O4    Data Outputs

**D.C. ELECTRICAL CHARACTERISTICS (Operating)** HM-7620/21-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-7620/21-2/8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/enable "1"	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	-	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	-	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4*	3.2*	-	V	$I_{OH} = -2.0mA$ , $V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA$ , $V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}$ , $V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40*	$\mu A$	$V_{OL} = 0.3V$ , $V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15*	-	-100*	mA	$V_{CC} = V_{CC} \text{ Max.}$ , $V_{OUT} = 0.0V$ One Output Only for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	90	120	mA	$V_{CC} = V_{CC} \text{ Max.}$ All Inputs Grounded

NOTE: Positive current defined as into device terminals.

\* "Three-State" only.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7620/21 - 5 5V $\pm 5\%$ 0°C to +75°C						HM-7620/21-2/8 5V $\pm 10\%$ -55°C to +125°C						UNITS
		"B"		"A"		STD		"B"		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	30	40	40	50	50	70	-	55	-	70	-	85	ns
T <sub>EA</sub>	Chip Enable Access Time	-	25	-	25	-	25	-	30	-	30	-	30	ns
T <sub>DA</sub>	Chip Disable Access Time	-	25	-	25	-	25	-	30	-	30	-	30	ns

A.C. limits guaranteed for worst case N2 sequencing with maximum test frequency of 5MHz.

# HM-7640/41

## HIGH SPEED 512 x 8 PROM

HM-7640 - Open Collector Outputs  
 HM-7641 - "Three State" Outputs

### Description

The HM-7640/41 are fully decoded high speed Schottky TTL 4096 bit Field Programmable ROMs in a 512 word by 8 bit/word format with open collector (HM-7640) or "three-state" (HM-7641) outputs. These PROMs are available in a 24 pin DIP (ceramic or power plastic).

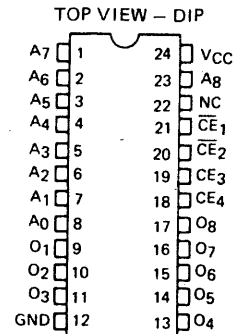
All bits are manufactured storing a logical "1" (positive logic) and can be selectively programmed for a logical "0" in any bit position.

Nickel-chromium fuse technology is used on this and all other Harris Bipolar PROMs.

The HM-7640/41 contain test rows and columns which are in addition to the storage array to assure high programmability and guarantee parametric and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

There are four chip enable inputs on the HM-7640/41 where  $\overline{CE}_1$  and  $\overline{CE}_2$  low and  $CE_3$  and  $CE_4$  high enables the chip.

### Pinouts



PIN NAMES

- A<sub>0</sub> - A<sub>8</sub> Address Inputs
- O<sub>1</sub> - O<sub>8</sub> Data Outputs
- $\overline{CE}_1, \overline{CE}_2, CE_3, CE_4$  Chip Enable Inputs

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-7640/41-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-7640/41-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	-40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	-	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	-	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4*	3.2*	-	V	$I_{OH} = -2.0mA, V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50/0.45	V	$I_{OL} = +16mA, V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}, V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40*	$\mu A$	$V_{OL} = 0.3V, V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15*	-	-100*	mA	$V_{CC} = V_{CC} \text{ Max.}, V_{OUT} = 0.0V$ One Output at a Time for Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	125	170	mA	$V_{CC} = V_{CC} \text{ Max.},$ All Inputs Grounded.

NOTE: Positive current defined as into device terminals.  
 \* "Three-State" only.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7640/41-5 5V $\pm 5\%$ 0°C to +75°C				HM-7640/41-2/-8 5V $\pm 10\%$ -55°C to +125°C				UNITS
		"A"		STD		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	35	45	50	70	-	60	-	85	ns
T <sub>EA</sub>	Chip Enable Access Time	-	35	-	40	-	45	-	50	ns
T <sub>DA</sub>	Chip Disable Access Time	-	35	-	40	-	45	-	50	ns

A.C. limits guaranteed for worst case N2 sequencing with maximum test frequency of 5MHz.

### Description

The HM-7649 is a fully decoded high speed Schottky TTL 4096 bit Field Programmable ROM in a 512 word by 8 bit/word format with "Three State" outputs. This PROM is available in a 20 pin D.I.P. (ceramic or power plastic).

All bits are manufactured storing a logical "1" (positive logic) and can be selectively programmed for a logical "0" in any bit position.

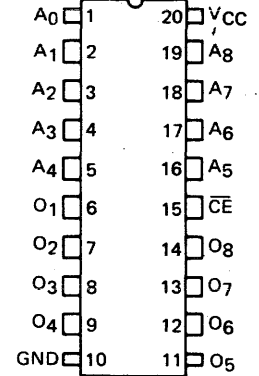
Nickel Chromium fuse technology is used on this and all other Harris Bipolar PROMs.

The pinout is identical to the 74S473 PROM.

The HM-7649 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parametric and A. C. performance. The fuses in these test rows and columns are blown prior to shipment.

There is a chip enable input on the HM-7649 where  $\overline{CE}$  low enables the device.

### Pinout



TOP VIEW - D.I.P.

PIN NAMES  
A<sub>0</sub>-A<sub>8</sub> Address Inputs  
O<sub>1</sub>-O<sub>8</sub> Data Outputs  
 $\overline{CE}$  Chip Enable Input

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-7649-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )

HM-7649-2/8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )

Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	+25	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100/-250	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	-	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Input Voltage "0"	-	-	0.80	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4	3.2	-	V	$I_{OH} = -2.0mA, V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Output Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA, V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}, V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Output Current "0"	-	-	-40	$\mu A$	$V_{OL} = 0.3V, V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-20	-	-100	mA	$V_{CC} = V_{CC} \text{ Max.}, V_{OUT} = 0.0V,$ One Output Only for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	120	170	mA	$V_{CC} = V_{CC} \text{ Max.}$ All Inputs Grounded

NOTE: Multiple entries refer to parameter values for "A"/STD. Positive current defined as into device terminals.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7649-5 5V $\pm 5\%$ 0°C to +75°C				HM-7649-2/-8 5V $\pm 10\%$ -55°C to +125°C				UNITS
		"A"		STD		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	35	45	40	60	-	60	-	80	ns
T <sub>EA</sub>	Chip Enable Access Time	-	35	-	40	-	45	-	50	ns
T <sub>DA</sub>	Chip Disable Access Time	-	35	-	40	-	45	-	50	ns

A.C. limits guaranteed for worst case N2 sequencing with maximum test frequency of 5MHz.

#### Description

The HM-7642/43 are fully decoded high speed Schottky TTL 4096 bit Field Programmable ROMs in a 1K word by 4 bit/word format with open collector (HM-7642) or "three state" (HM-7643) outputs. These PROM's are available in an 18 pin DIP (ceramic or power plastic).

All bits are manufactured storing a logical "1" (positive logic) and can be selectively programmed for a logical "0" in any bit position.

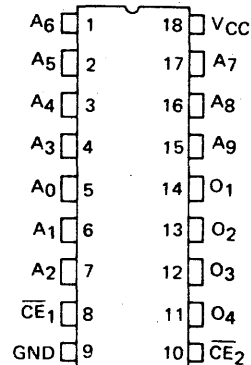
Nickel-chromium fuse technology is used on these and all other Harris Bipolar PROMs.

The HM-7642/43 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parameters and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

There are two chip enable inputs on the HM-7642/43.  $\overline{CE}_1$  and  $\overline{CE}_2$  low enables the chip.

#### Pinout

TOP VIEW-DIP



PIN NAMES

- A<sub>0</sub> - A<sub>9</sub> ADDRESS INPUTS
- O<sub>1</sub> - O<sub>4</sub> DATA OUTPUTS
- $\overline{CE}_1$ ,  $\overline{CE}_2$  CHIP ENABLE INPUTS

#### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-7642/43-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )

HM-7642/43-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
$I_{IH}$ $I_{IL}$	Address/Enable "1" Input Current "0"	-	-	-40 -100	$\mu A$ $\mu A$	$V_{IH} = V_{CC} \text{ Max.}$ $V_{IL} = 0.45V$
$V_{IH}$ $V_{IL}$	Input Threshold "1" Voltage "0"	2.0 -	- -	- 0.8	V V	$V_{CC} = V_{CC} \text{ Min.}$ $V_{CC} = V_{CC} \text{ Max.}$
$V_{OH}$ $V_{OL}$	Output "1" Voltage "0"	2.4*	3.2*	- 0.50/0.50/0.45	V V	$I_{OH} = -2.0mA$ , $V_{CC} = V_{CC} \text{ Min.}$ $I_{OL} = +16mA$ , $V_{CC} = V_{CC} \text{ Min.}$
$I_{OHE}$ $I_{OLE}$	Output Disable "1" Current "0"	-	-	+40 -40*	$\mu A$ $\mu A$	$V_{OH}$ , $V_{CC} = V_{CC} \text{ Max.}$ $V_{OL} = 0.3V$ , $V_{CC} = V_{CC} \text{ Max.}$
$V_{CL}$	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
$I_{OS}$	Output Short Circuit Current	-15*	-	-100*	mA	$V_{CC} = V_{CC} \text{ Max.}$ , $V_{OUT} = 0.0V$ , One Output at a Time for a Max. of 1 Second
$I_{CC}$	Power Supply Current	-	100	140	mA	$V_{CC} = V_{CC} \text{ Max.}$ , All Inputs Grounded.

NOTE: Positive current defined as into device terminals.

\* "Three-State" only.

#### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7642/43-5 5V $\pm 5\%$ 0°C to +75°C						HM-7642/43-2/-8 5V $\pm 10\%$ -55°C to +125°C						UNITS
		"B"		"A"		STD		"B"		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
TAA	Address Access Time	35	45	40	50	45	60	-	55	-	70	-	85	ns
TEA	Chip Enable Access Time	-	25	-	25	-	25	-	30	-	30	-	30	ns
TDA	Chip Disable Access Time	-	25	-	25	-	25	-	30	-	30	-	30	ns

A.C. limits guaranteed for worst case N2 sequencing with maximum test frequency of 5MHz.



### Description

The HM-7681 is a fully decoded high speed Schottky-TTL 8192 bit Field Programmable ROM in a 1K word by 8 bit/word format with "Three State" outputs. This PROM is available in a 24 pin D.I.P. (ceramic or power plastic).

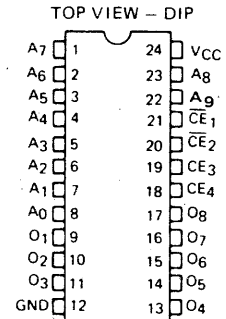
All bits are manufactured storing a logic "1" (Positive Logic) and can be selectively programmed for a logical "0" in any one bit position.

Nickel-chromium fuse technology is used on this and all other HARRIS Bipolar PROMs.

The HM-7681 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parametric and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

There are four chip enable inputs on the HM-7681.  $\overline{CE}_1$ ,  $\overline{CE}_2$  low, and  $\overline{CE}_3$ ,  $\overline{CE}_4$  high enables the chip.

### Pinouts



#### PIN NAMES

- A<sub>0</sub> - A<sub>7</sub> Address Inputs
- O<sub>1</sub> - O<sub>8</sub> Data Outputs
- $\overline{CE}_1$ ,  $\overline{CE}_2$ ,  $\overline{CE}_3$ ,  $\overline{CE}_4$  Chip Enable Inputs

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-7681-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-7681-2/8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/enable "1" Input Current	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100/-250	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold Voltage "1"	2.0	-	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Input Threshold Voltage "0"	-	-	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output Voltage "1"	2.4	3.2	-	V	$I_{OH} = -2.0mA$ , $V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Output Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA$ , $V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1" Current	-	-	+40	$\mu A$	$V_{OH}$ , $V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Output Disable "0" Current	-	-	-40	$\mu A$	$V_{OL} = 0.3V$ , $V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15	-	-100	mA	$V_{CC} = V_{CC} \text{ Max.}$ , $V_{OUT} = 0.0V$ One Output Only for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	130	170	mA	$V_{CC} = V_{CC} \text{ Max.}$ All Inputs Grounded

NOTE: Multiple entries refer to parameter values for "A"/STD. Positive current defined as into device terminals.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7681-5 5V $\pm 5\%$ 0°C to +75°C				HM-7681-2/8 5V $\pm 10\%$ -55°C to +125°C				UNITS
		"A"		STD		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	40	50	50	70	-	60	-	90	ns
T <sub>EA</sub>	Chip Enable Access Time	-	35	-	40	-	40	-	50	ns
T <sub>DA</sub>	Chip Disable Access Time	-	35	-	40	-	40	-	50	ns

A.C. Limits guaranteed for worst case N<sup>2</sup> sequencing with maximum test frequency of 5MHz.

### Description

The HM-7685 is a fully decoded high speed Schottky TTL 8192-bit Field Programmable ROM in a 2K word by a 4 bit/word format with "Three State" outputs. This PROM is available in an 18 pin DIP (ceramic or power plastic).

All bits are manufactured storing a logical "1" (positive logic) and can be selectively programmed for a logical "0" in any bit position.

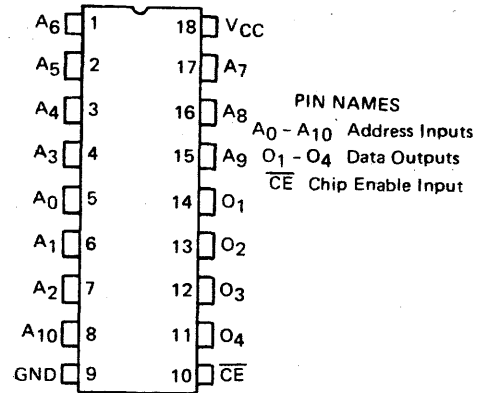
Nickel-chromium fuse technology is used on this and all other Harris Bipolar PROMs.

The HM-7685 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parameters and A. C. performance. The fuses in these test rows and columns are blown prior to shipment.

There is a chip enable on the HM-7685.  $\overline{CE}$  low enables the chip.

### Pinouts

TOP VIEW - DIP



D.C. ELECTRICAL CHARACTERISTICS (Operating) HM-7685-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-7685-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100/-250	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	-	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	-	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4	3.2	-	V	$I_{OH} = -2.0mA, V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA, V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}, V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40	$\mu A$	$V_{OL} = 0.3V, V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15	-	-100	mA	$V_{CC} = V_{CC} \text{ Max.}, V_{OUT} = 0.0V,$ One Output at a Time for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	120	170	mA	$V_{CC} = V_{CC} \text{ Max.},$ All Inputs Grounded.

NOTE: Multiple entries refer to parameter values of "A"/STD. Positive current defined as into device terminals.

A.C. ELECTRICAL CHARACTERISTICS (Operating)  
 Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-7685-5 5V $\pm 5\%$ 0°C to +75°C				HM-7685-2/-8 5V $\pm 10\%$ -55°C to +125°C				UNITS
		"A"		STD		"A"		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	40	50	50	70	-	60	-	90	ns
T <sub>EA</sub>	Chip Enable Access Time	-	30	-	40	-	35	-	50	ns
T <sub>DA</sub>	Chip Disable Access Time	-	30	-	40	-	35	-	50	ns

A.C. limits guaranteed for worst case N2 sequencing with maximum test frequency of 5MHz.

Harris Semiconductor  
MEMORY

### Description

The HM-76161 is a fully decoded high speed Schottky TTL 16,384 bit Field Programmable ROM in a 2K word by 8 bit/word format with "Three State" outputs. This PROM is available in a 24 pin DIP (ceramic or power plastic).

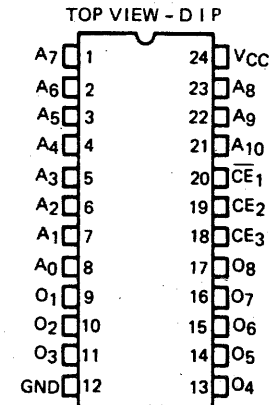
All bits are manufactured storing a logical "1" (Positive Logic) and can be selectively programmed for a logical "0" in any bit position.

The nickel-chromium fuse technology used is the same as all other Harris Bipolar PROMs and the JAN approved MIL-M-38510 PROMs.

The HM-76161 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parameters and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

There are three chip enable inputs on the HM-76161.  $\overline{CE}_1$  low,  $CE_2$  high, and  $CE_3$  high enables the device.

### Pinout



#### PIN NAMES

- A<sub>0</sub> - A<sub>10</sub> Address Inputs
- O<sub>1</sub> - O<sub>8</sub> Data Outputs
- $\overline{CE}_1$ ,  $CE_2$ ,  $CE_3$  Chip Enable Inputs

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-76161-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )

HM-76161-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-	-100	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	-	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	-	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4	3.2	-	V	$I_{OH} = -2.0mA$ , $V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA$ , $V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}$ , $V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40	$\mu A$	$V_{OL} = 0.3V$ , $V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15	-	-100	mA	$V_{CC} = V_{CC} \text{ Max.}$ $V_{OUT} = 0.0V$ One Output at a Time for Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	130	180	mA	$V_{CC} = V_{CC} \text{ Max.}$ , All Inputs Grounded.

NOTE: Positive current defined as into device terminals.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-76161-5 5V $\pm 5\%$ 0°C to +75°C				HM-76161-2/-8 5V $\pm 10\%$ -55°C to +125°C		UNITS
		"A"		STD		STD		
		TYP	MAX	TYP	MAX	TYP	MAX	
T <sub>AA</sub>	Address Access Time	40	50	45	60	-	80	ns
T <sub>EA</sub>	Chip Enable Access Time	-	40	-	40	-	50	ns
T <sub>DA</sub>	Chip Disable Access Time	-	40	-	40	-	50	ns

A.C. limits guaranteed for worst case N<sup>2</sup> sequencing with maximum test frequency of 5MHz

### Description

The HM-76165 PROM is a fully decoded Schottky TTL 16,384 bit field programmable ROM in a 4K word by 4 bit/word format with "Three State" outputs. This PROM is available in a 20 pin 0.300 inch wide DIP.

All bits are manufactured storing a logical "1" and can be selectively programmed for a logical "0" in any bit position.

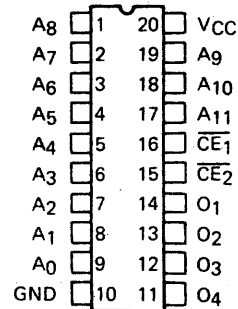
MIL-M-38510 qualified NiCr technology is used on the HM-76165 and all other HARRIS bipolar PROMs.

The HM-76165 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parametrics and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

The HM-76165 utilizes two chip enables where  $\overline{CE}_1$  low and  $\overline{CE}_2$  low enables the device.

### Pinout

TOP VIEW - DIP



PIN NAMES

- A<sub>0</sub> - A<sub>11</sub> Address Inputs
- O<sub>1</sub> - O<sub>4</sub> Data Outputs
- $\overline{CE}_1, \overline{CE}_2$  Chip Enable Inputs

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-76165-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-76165-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-50.0	-100	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	1.5	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	1.5	0.80	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4	3.2	-	V	$I_{OH} = -2.0mA, V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA, V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE(1)</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}, V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40	$\mu A$	$V_{OL} = 0.3V, V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15	-	-100	mA	$V_{OUT} = 0.0V$ , One Output at a Time for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	120	170	mA	$V_{CC} = V_{CC} \text{ Max.}$ , All Inputs Grounded.

NOTE: Positive current defined as into device terminals.  
 (1) I<sub>OHE</sub> = +60 $\mu A$  for -2 and -8.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = +125^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-76165-5 5V $\pm 5\%$ 0°C to +75°C		HM-76165-2/-8 5V $\pm 10\%$ -55°C to +125°C		UNITS
		TYPICAL	MAXIMUM	TYPICAL	MAXIMUM	
T <sub>AA</sub>	Address Access Time	45	60	50	80	ns
T <sub>EA</sub>	Chip Enable Access Time	25	35	30	40	ns
T <sub>DA</sub>	Chip Disable Access Time	25	35	30	40	ns

A.C. limits guaranteed for worst case N<sup>2</sup> sequencing with maximum test frequency of 5MHz.

### Description

The HM-76321 is a fully decoded Schottky TTL 32,768 bit field programmable ROM in a 4K word by 8 bit/word format with "Three State" outputs. This PROM is available in a 24 pin DIP.

All bits are manufactured storing a logical "1" (Positive Logic) and can be selectively programmed for a logical "0" in any bit position.

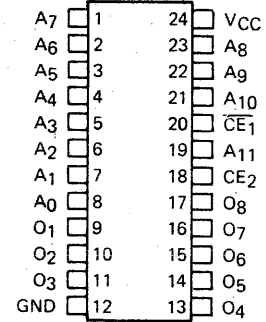
MIL-M-38510 qualified NiCr technology is used on the HM-76321 and all other HARRIS Bipolar PROMs.

The HM-76321 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parameters and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

The HM-76321 utilizes two chip enable controls, where  $\overline{CE}_1$  low and  $CE_2$  high enables the device.

### Pinout

TOP VIEW - DIP



PIN NAMES

A<sub>0</sub> - A<sub>11</sub> Address Inputs  
 O<sub>1</sub> - O<sub>8</sub> Data Outputs  
 $\overline{CE}_1$ , CE<sub>2</sub> Chip Enable Inputs

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-76321-5 (V<sub>CC</sub> = 5.0V ±5%, T<sub>A</sub> = 0°C to +75°C)  
 HM-76321-2/-8 (V<sub>CC</sub> = 5.0V ±10%, T<sub>A</sub> = -55°C to +125°C)  
 Typical Measurements are at T<sub>A</sub> = 25°C, V<sub>CC</sub> = +5V

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub> I <sub>IL</sub>	Address/Enable "1" Input Current "0"	-	-	+40 -100	μA	V <sub>IH</sub> = V <sub>CC</sub> Max. V <sub>IL</sub> = 0.45V
V <sub>IH</sub> V <sub>IL</sub>	Input Threshold "1" Voltage "0"	2.0	1.5	- 0.8	V	V <sub>CC</sub> = V <sub>CC</sub> Min. V <sub>CC</sub> = V <sub>CC</sub> Max.
V <sub>OH</sub> V <sub>OL</sub>	Output "1" Voltage "0"	2.4	3.2	- 0.50	V	I <sub>OH</sub> = -2.0mA, V <sub>CC</sub> = V <sub>CC</sub> Min. I <sub>OL</sub> = +16mA, V <sub>CC</sub> = V <sub>CC</sub> Min.
I <sub>OHE(1)</sub> I <sub>OLE</sub>	Output Disable "1" Current "0"	-	-	+40 -40	μA	V <sub>OH</sub> , V <sub>CC</sub> = V <sub>CC</sub> Max. V <sub>OL</sub> = 0.3V, V <sub>CC</sub> = V <sub>CC</sub> Max.
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	I <sub>IN</sub> = -18mA
I <sub>OS</sub>	Output Short Circuit Current	-15	-	-100	mA	V <sub>OUT</sub> = 0.0V, One Output at a Time for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	-	190	mA	V <sub>CC</sub> = V <sub>CC</sub> Max., All Inputs Grounded.

NOTE: Positive current defined as into device terminals.  
 (1) I<sub>OHE</sub> = +100μA for -2 and -8.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical measurements are at T<sub>A</sub> = +25°C, V<sub>CC</sub> = +5V

SYMBOL	PARAMETER	HM-76321-5 5V ±5% 0°C to +75°C		HM-76321-2/-8 5V ±10% -55°C to +125°C		UNITS
		TYPICAL	MAXIMUM	TYPICAL	MAXIMUM	
T <sub>AA</sub>	Address Access Time	45	65	-	85	ns
T <sub>EA</sub>	Chip Enable Access Time	25	35	-	40	ns
T <sub>DA</sub>	Chip Disable Access Time	25	35	-	40	ns

A.C. limits guaranteed for worst case N<sup>2</sup> sequencing with maximum test frequency of 5MHz.

### Description

The HM-76641 is a fully decoded Schottky TTL 65,536 bit field programmable ROM in an 8K word by 8 bit/word format with "Three State" outputs. This PROM is available in a 24 pin DIP.

All bits are manufactured storing a logical "1" (Positive Logic) and can be selectively programmed for a logical "0" in any bit position.

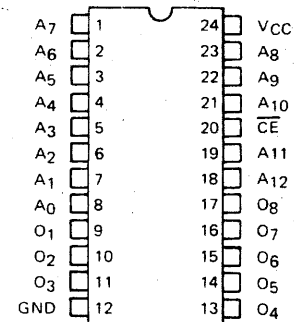
MIL-M-38510 qualified NiCr technology is used on the HM-76641 and all other HARRIS bipolar PROMs.

The HM-76641 contains test rows and columns which are in addition to the storage array to assure high programmability and guarantee parameters and A.C. performance. The fuses in these test rows and columns are blown prior to shipment.

The HM-76641 utilizes a single chip enable,  $\overline{CE}$ , which when low enables the device.

### Pinout

TOP VIEW - DIP



PIN NAMES

A<sub>0</sub> - A<sub>12</sub> Address Inputs  
 O<sub>1</sub> - O<sub>8</sub> Data Outputs  
 $\overline{CE}$  Chip Enable Inputs

### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HM-76641-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HM-76641-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )  
 Typical Measurements are at  $T_A = 25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
I <sub>IH</sub>	Address/Enable "1"	-	-	+40	$\mu A$	$V_{IH} = V_{CC} \text{ Max.}$
I <sub>IL</sub>	Input Current "0"	-	-50.0	-100	$\mu A$	$V_{IL} = 0.45V$
V <sub>IH</sub>	Input Threshold "1"	2.0	1.5	-	V	$V_{CC} = V_{CC} \text{ Min.}$
V <sub>IL</sub>	Voltage "0"	-	1.5	0.8	V	$V_{CC} = V_{CC} \text{ Max.}$
V <sub>OH</sub>	Output "1"	2.4	3.2	-	V	$I_{OH} = -2.0mA$ , $V_{CC} = V_{CC} \text{ Min.}$
V <sub>OL</sub>	Voltage "0"	-	0.35	0.50	V	$I_{OL} = +16mA$ , $V_{CC} = V_{CC} \text{ Min.}$
I <sub>OHE(1)</sub>	Output Disable "1"	-	-	+40	$\mu A$	$V_{OH}$ , $V_{CC} = V_{CC} \text{ Max.}$
I <sub>OLE</sub>	Current "0"	-	-	-40	$\mu A$	$V_{OL} = 0.3V$ , $V_{CC} = V_{CC} \text{ Max.}$
V <sub>CL</sub>	Input Clamp Voltage	-	-	-1.2	V	$I_{IN} = -18mA$
I <sub>OS</sub>	Output Short Circuit Current	-15	-	-100	mA	$V_{OUT} = 0.0V$ , One Output at a Time for a Max. of 1 Second
I <sub>CC</sub>	Power Supply Current	-	-	190	mA	$V_{CC} = V_{CC} \text{ Max.}$ , All Inputs Grounded.

NOTE: Positive current defined as into device terminals.

(1) I<sub>OHE</sub> = +100 $\mu A$  for -2 and -8.

### A.C. ELECTRICAL CHARACTERISTICS (Operating)

Typical Measurements are at  $T_A = +25^\circ C$ ,  $V_{CC} = +5V$

SYMBOL	PARAMETER	HM-76641-5 5V $\pm 5\%$ 0 $^\circ C$ to +75 $^\circ C$		HM-76641-2/-8 5V $\pm 10\%$ -55 $^\circ C$ to +125 $^\circ C$		UNITS
		TYPICAL	MAXIMUM	TYPICAL	MAXIMUM	
T <sub>AA</sub>	Address Access Time	50	85	-	100	ns
T <sub>EA</sub>	Chip Enable Access Time	30	40	-	45	ns
T <sub>DA</sub>	Chip Disable Access Time	30	40	-	45	ns

A.C. limits guaranteed for worst case N<sup>2</sup> sequencing with maximum test frequency of 5MHz.

## PROGRAMMING THE HM-76XXX PROMS

The HM-76XXX PROMs are manufactured with all bits storing a logical "1" (output high). Any desired bit can be programmed to a logical "0" (output low) by following the simple procedure shown below. One may build his own programmer to satisfy the specifications described in the table, or use any of the commercially available programmers which can meet these specifications.

### PROGRAMMING SPECIFICATIONS

SYMBOL	PARAMETER	MINIMUM	RECOMMENDED OR TYPICAL	MAXIMUM	UNITS
V <sub>IH</sub> V <sub>IL</sub>	Address Input Voltage (1)	2.4 0.0	5.0 0.4	5.0 0.5	V V
V <sub>PH</sub> (2) V <sub>PL</sub> (3)	Programming/Verify Voltage to V <sub>CC</sub>	12.0 4.5	12.0 4.5	12.5 5.5	V V
I <sub>IIP</sub>	Programming Input Low Current at V <sub>PH</sub>	—	-300	-600	μA
t <sub>r</sub> t <sub>f</sub>	Programming (V <sub>CC</sub> ) Voltage Rise and Fall Time	1.0 1.0	1.0 1.0	10.0 10.0	μs μs
t <sub>d</sub>	Programming Delay	10	10	100	μs
t <sub>p</sub>	Programming Pulse Width (4)	90	100	110	μs
P.D.C.	Programming Duty Cycle	—	50	90	%
V <sub>OPe</sub> V <sub>OPD</sub>	Output Voltage — Enable (5) Output Voltage — Disable (6)	10.5 4.5	10.5 5.0	11.0 5.5	V V
T <sub>a</sub>	Ambient Temperature	—	25	75	°C

During programming the chip must be disabled for proper operation.

- NOTES: 1. No inputs should be left open for V<sub>IH</sub>.  
 2. V<sub>PH</sub> source must be capable of supplying one ampere.  
 3. It is recommended that dual verification be made at V<sub>PL</sub> min and V<sub>PL</sub> max.  
 4. Note step 10 in programming procedure.  
 5. V<sub>OPe</sub> source must be capable of supplying 10mA minimum.  
 6. Disable condition will be met with output open circuit.

### PROGRAMMING PROCEDURE

- Address the PROM with the binary address of the word to be programmed. Address inputs are TTL compatible. An open circuit should not be used to address the PROM.
- Bring the  $\overline{CE}_x$  input(s) high and the CE<sub>x</sub> input(s) low to disable the device. The disabling of the device during programming is an essential step in correctly programming all HARRIS PROMs. The chip enables are TTL compatible. An open circuit should not be used to disable the device.
- Disable the programming circuitry by applying a voltage of V<sub>OPD</sub> to the outputs of the PROM. Any output may be left open to achieve the disable.
- Raise V<sub>CC</sub> to V<sub>PH</sub> with rise time less than or equal to t<sub>r</sub>.
- After a delay equal to or greater than t<sub>d</sub>, apply a pulse with amplitude of V<sub>OPe</sub> and duration of t<sub>p</sub> to the output selected for programming. Note that the PROM is supplied with fuses intact, which generates an output high. Programming a fuse will cause the output to go low.
- Other bits in the same word may be programmed while the V<sub>CC</sub> input is raised to V<sub>PH</sub> by applying output enable pulses to each output which is to be programmed. The output enable pulses must be separated by a minimum interval of t<sub>d</sub>.
- Lower V<sub>CC</sub> to 4.5 volts following a delay of t<sub>d</sub> from last programming enable pulse applied to an output.
- Enable the PROM for verification by applying V<sub>IL</sub> to  $\overline{CE}_x$  and V<sub>IH</sub> to CE<sub>x</sub>.
- Repeat verification (step 8) at V<sub>CC</sub> = 5.5 volts.
- If any bit does not verify as programmed, repeat steps 2 through 9 until the bit has received a total of 1ms of programming time. Bits which do not program within 1ms are programming rejects. No further attempt to program these parts should be made.
- Repeat steps 1 through 10 for all other bits to be programmed in the PROM.
- Programming rejects returned to the factory must be accompanied by data giving address, desired data, and actual output data of the location in which a programming failure has occurred.



### TYPICAL PROGRAMMING CIRCUIT

The circuit and timing diagrams shown in Figures 1 and 2 will establish the proper programming conditions for the output enable pulses. This allows the use of standard TTL

parts for all logic inputs to the PROM. Note the gate which senses the output must withstand inputs up to 11.0 volts during programming.

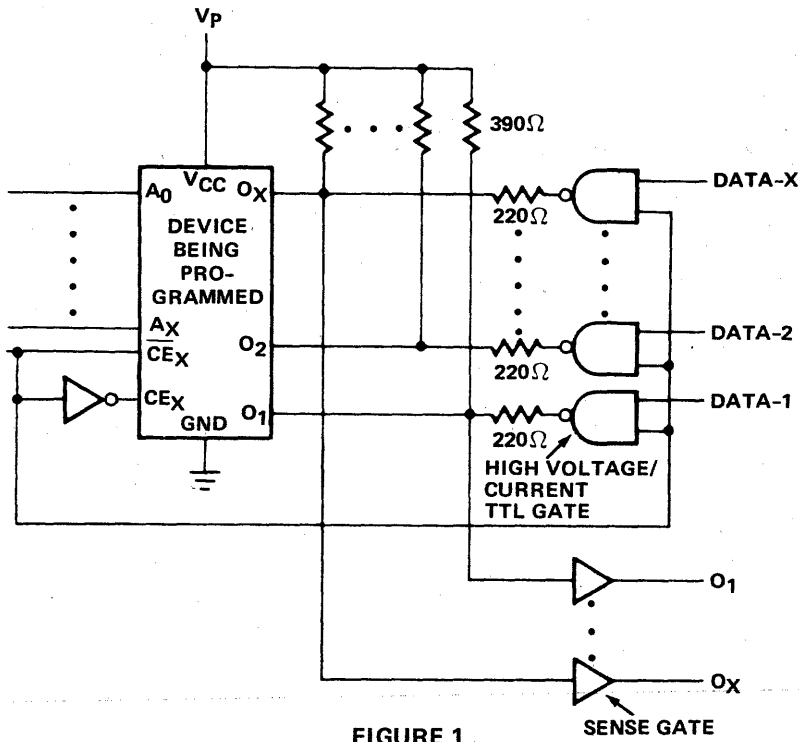


FIGURE 1

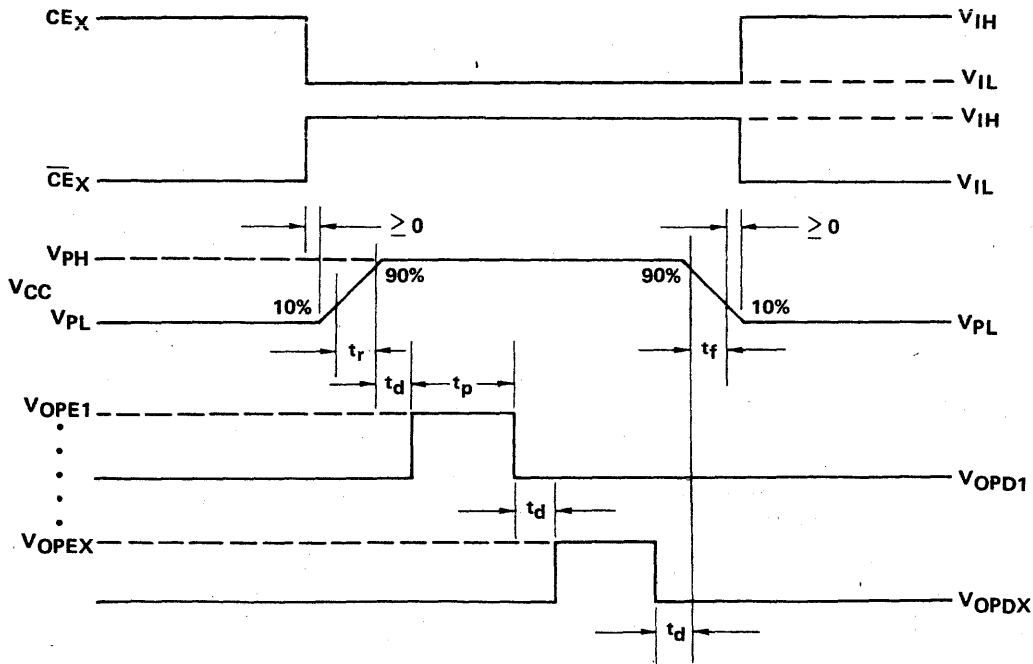


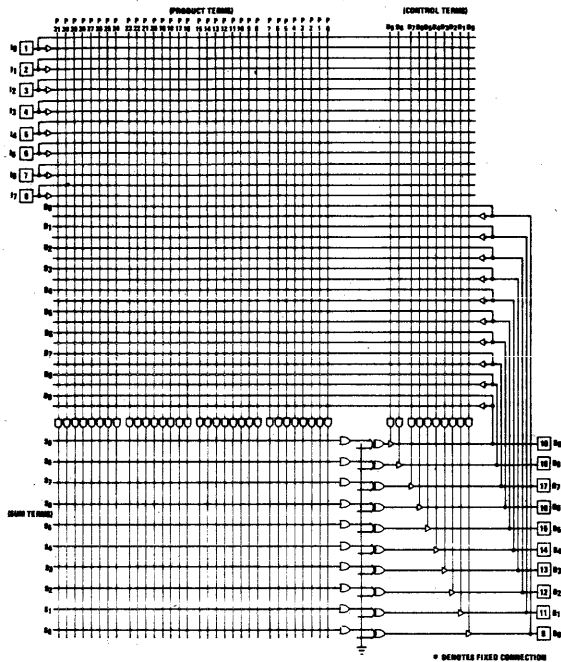
FIGURE 2

Waveforms Applied to the Device Pins During Programming

MEMORY Harris Semiconductor



### Functional Diagram



### D. C. ELECTRICAL CHARACTERISTICS (Operating)

HPL-77153/82S153-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HPL-77153/82S153-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )

SYMBOL	PARAMETER	MIN	MAX	UNITS	TEST CONDITIONS
$I_{IH}$ $I_{IL}$	Input Current Dedicated Input	"1" "0"	- -100	$\mu A$	$V_{IH} = V_{CC} \text{ MAX}$ $V_{IL} = 0.4V$ $V_{CC} = V_{CC} \text{ MAX}$
$I_{BZH}$ $I_{BZL}$	Input Current Bidirectional Pin Current Hi-Z State	"1" "0"	- -100	$\mu A$	$V_{BH} = V_{CC} \text{ MAX}$ $V_{BL} = 0.4V$ $V_{CC} = V_{CC} \text{ MAX}$
$V_{IH}$ $V_{IL}$	Input Threshold Voltage	"1" "0"	2.0 0.8	V	$V_{CC} = V_{CC} \text{ MAX}$ $V_{CC} = V_{CC} \text{ MIN}$
$V_{OH}$ $V_{OL}$	High-Level Output Voltage Low-Level Output Voltage	2.4 0.5	- 0.5	V	$V_{CC} = V_{CC} \text{ MIN}$ $V_{IL} = 0.8V$ $V_{IH} = 2.0V$ $I_{OH} = -2.0mA$ $I_{OL} = +16mA$
$V_{CL}$	Input Clamp Voltage	-	-1.2	V	$I_{IN} = -18mA, V_{CC} = 0V$
$I_{OS}$	Output Short Circuit Current	C M	-20 -85	mA	$V_{CC} = 5.0V, V_{OUT} = 0V$ One output for MAX of One Sec.
$I_{CC}$	Power Supply Current	C M	- 165	mA	$V_{CC} = V_{CC} \text{ MAX}$

C = Commercial (-5) M = Military (-2/-8)

### A. C. SWITCHING CHARACTERISTICS (Operating)

SYMBOL	JEDEC STANDARD	OLD SYMBOL	PARAMETER	HPL-77153/82S153-5 5V $\pm$ 5% 0°C to +75°C		HPL-77153/82S153-2/-8 5V $\pm$ 10% -55°C to +125°C		UNITS
				MIN	MAX	MIN	MAX	
TDVQH1		TPD	Propagation Delay-Input or I/O to Active High Output	-	40	-	55	ns
TDVQL1		TPD	Propagation Delay-Input or I/O to Active Low Output	-	40	-	55	ns
TDVQH2		TOE	Enable Access Time to Active High Output	-	35	-	45	ns
TDVQL2		TOE	Enable Access Time to Active Low Output	-	35	-	45	ns
TDVQZ1		TOD	Disable Access Time from Active High Output	-	30	-	45	ns
TDVQZ2		TOD	Disable Access Time from Active Low Output	-	30	-	45	ns

NOTE: Maximum test frequency is 5MHz with a 50% duty cycle.

### Description

The HPL-77153/82S153 is a programmable logic device designed to be cost effective and space saving replacement for discrete logic designs. This device is a two-level logic element, consisting of 32 product terms (AND) and 10 sum terms (OR) with fusible links for programming I/O polarity and direction.

All product terms can be linked to 8 inputs (I) and 10 bidirectional I/O lines (B) allowing variable I/O configurations using the 10 direction control gates (D), ranging from 17 inputs and 1 output to 8 inputs and 10 outputs.

On chip T/C buffers allow either True (I,B) and/or Complement ( $\bar{I}, \bar{B}$ ) signals to be linked to any of the product terms, the outputs of which may be used as inputs to any or all of the sum terms. The output polarity of the sum terms is individually program-

mable by means of a fuse link controlled EX-OR gate to allow implementation of Sum Of Products (SOP) or Inverted Sum Of Products (ISOP) expressions.

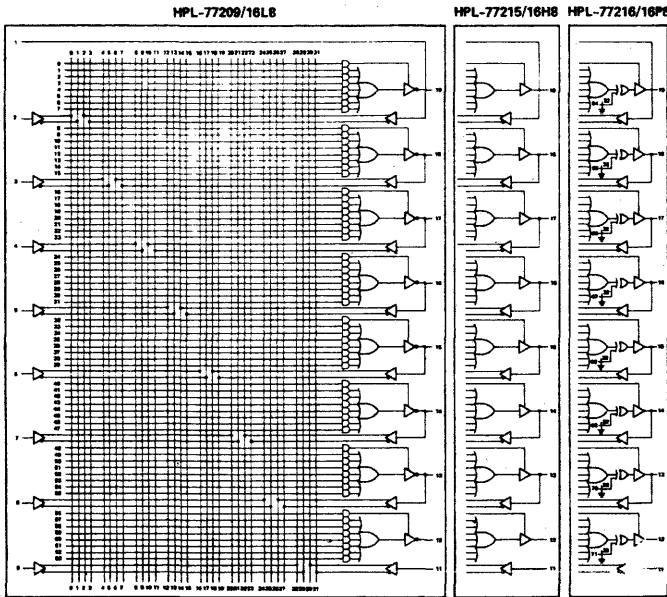
The HPL-77153/82S153 is field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

Nickel-chromium fuse technology is used on these and all other HARRIS HPL-77XXX programmable logic devices. The HPL-77153/82S153 is available in 20 pin slimline DIP packages with a pinout identical to the 82S153.

The HPL-77153/82S153 contains unique test circuitry developed by HARRIS which is enabled at the time of manufacture to allow complete AC and DC testing.

## HPL™ Harris Programmable Logic

### Functional Diagram



#### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HPL-77215/216/209 - 5 (V<sub>CC</sub> = 5.0V ± 5%, T<sub>A</sub> = 0°C to +75°C)  
HPL-77215/216/209 - 2/-8 (V<sub>CC</sub> = 5.0V ± 10%, T<sub>A</sub> = -55°C to +125°C)

SYMBOL	PARAMETER	MIN	MAX	UNITS	TEST CONDITIONS
I <sub>ih</sub>	Input Current	-1*	+25	μA	V <sub>ih</sub> = V <sub>CC</sub> MAX
I <sub>ic</sub>	Dedicated Input	0*	-100	μA	V <sub>ih</sub> = 0.4 V
I <sub>oh</sub>	Output Current	-1*	+40	μA	V <sub>oh</sub> = V <sub>CC</sub> MAX
I <sub>oz</sub>	High-Z State	0*	-40	μA	V <sub>ih</sub> = 0.4 V
I <sub>bi</sub>	Bidirectional Pin Current	-1*	-40	μA	V <sub>ih</sub> = V <sub>CC</sub> MAX
I <sub>bz</sub>	High-Z State	0*	-100	μA	V <sub>ih</sub> = 0.4 V
V <sub>ih</sub>	Input Threshold (1)	2.0	-	V	V <sub>CC</sub> = V <sub>CC</sub> MAX
V <sub>il</sub>	Input Threshold (1)	0.8	-	V	V <sub>CC</sub> = V <sub>CC</sub> MIN
V <sub>oh</sub>	High-Level Output Voltage (2)	2.4	-	V	V <sub>CC</sub> = V <sub>CC</sub> MIN
V <sub>ol</sub>	Low-Level Output Voltage (2)	-	0.5	V	V <sub>ih</sub> = 0.8 V
V <sub>ic</sub>	Input Clamp Voltage (1)	-	-1.2	V	V <sub>ih</sub> = 2.0 V
I <sub>os</sub>	Output Short Circuit Current (2)	-30	-130	mA	V <sub>ih</sub> = -16mA, V <sub>CC</sub> = 0 V
I <sub>cc</sub>	Power Supply Current	-	155 (C)	mA	V <sub>CC</sub> = 5.0V, V <sub>oh</sub> = 0 V
		-	165 (M)	mA	One Output for MAX of One Sec.

C = Commercial (-5) M = Military (-2/-8)  
(1) These specifications apply to both Input (I) and Bidirectional (B) pins.  
(2) These specifications apply to both Output (F) and Bidirectional (B) pins.  
(3) One output at a time, otherwise 16mA.

#### A. C. SWITCHING CHARACTERISTICS (Operating)

SYMBOL	JEDEC STANDARD	OLD SYMBOL	PARAMETER	HPL-77215-5 HPL-77216-5 HPL-77209-5 0°C to +75°C		HPL-77215-2/-8 HPL-77216-2/-8 HPL-77209-2/-8 -55°C to +125°C		UNITS
				MIN	MAX	MIN	MAX	
TDVQH1		T <sub>PD</sub>	Propagation Delay - Input or I/O to Active High Output	-	35	-	45	ns
TDVQL1		T <sub>PD</sub>	Propagation Delay - Input or I/O to Active Low Output	-	35	-	45	ns
TDVQH2		T <sub>PRZ</sub>	Enable Access Time to Active High Output (1)	TDVQZ1	35	TDVQZ1	45	ns
TDVQL2		T <sub>PRZ</sub>	Enable Access Time to Active Low Output (1)	TDVQZ2	35	TDVQZ2	45	ns
TDVQZ1		T <sub>PRZ</sub>	Disable Access Time from Active High Output	-	30	-	35	ns
TDVQZ2		T <sub>PRZ</sub>	Disable Access Time from Active Low Output	-	30	-	35	ns

(1) Enable Access Time is guaranteed greater than Disable Access Time to avoid device contention.  
NOTE: Maximum test frequency is 50kHz with a 50% duty cycle.

### Description

The HPL-77215/216/209 are programmable logic devices designed to be cost effective and space saving replacements for discrete logic designs. These devices are two-level logic elements consisting of 7 product terms (AND) summed (OR) together to generate each of the 8 outputs. An eighth product term associated with each output can drive it to a high impedance state allowing 6 (B0-B5) of the 8 outputs to be used as inputs, either permanently or dynamically.

The HPL-77209 is functionally identical to the industry standard 16L8, and implements logic expressions of the Inverted Sum Of Products (ISOP) form.

The HPL-77215 is a similar device to the 16L8 but does not include the associated output inversion. It implements logic expressions of the Sum Of Products (SOP) form.

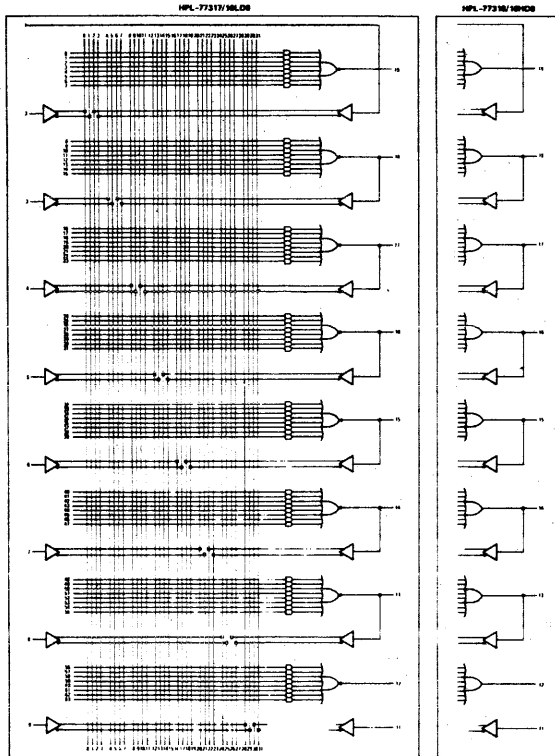
The HPL-77216 is a more flexible device, it includes 8 EX-OR gates in each output path, controlled by 8 extra fuses, allowing the polarity of each output to be user configured. This device can implement a combination of SOP and ISOP expressions.

The HPL-77215/216/209 are field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

Nickel-chromium fuse technology is used on these and all other HARRIS HPL-77XXX programmable logic devices. The HPL-77215/216/209 are available in 20 pin slimline DIP packages with pinouts identical to the 16L8.

The HPL-77215/216/209 contain unique test circuitry developed by HARRIS which is enabled at the time of manufacture to allow complete AC and DC testing.

### Functional Diagram



#### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HPL-77317/318 -5 (V<sub>CC</sub> = 5.0V±5%, T<sub>A</sub> = 0°C to +75°C)  
 HPL-77317/318 -2/-8 (V<sub>CC</sub> = 5.0V±10%, T<sub>A</sub> = -55°C to +125°C)

SYMBOL	PARAMETER	MIN	MAX	UNITS	TEST CONDITIONS
I <sub>in</sub>	Input Current	-1	-25	μA	V <sub>in</sub> = V <sub>CC</sub> MAX
I <sub>is</sub>	Dedicated Input	0	100	μA	V <sub>in</sub> = 0.4 V
V <sub>ih</sub>	Input Threshold Voltage	2.0	0.8	V	V <sub>CC</sub> = V <sub>CC</sub> MAX
V <sub>il</sub>	Input Threshold Voltage	0	0.8	V	V <sub>CC</sub> = V <sub>CC</sub> MIN
V <sub>oh</sub>	High-Level Output Voltage	2.4	-	V	V <sub>CC</sub> = V <sub>CC</sub> MIN, I <sub>OH</sub> = -2.0mA (M) I <sub>OH</sub> = -3.2mA (C)
V <sub>ol</sub>	Low-Level Output Voltage	-	0.5	V	V <sub>CC</sub> = 0.8V, I <sub>OL</sub> = +16mA (M) I <sub>OL</sub> = +24mA (C) (1)
V <sub>ci</sub>	Input Clamp Voltage (1)	-	-1.2	V	I <sub>in</sub> = -18mA, V <sub>CC</sub> = 0V
I <sub>os</sub>	Output Short Circuit Current	-30	-130	mA	V <sub>CC</sub> = 5.0V, V <sub>OUT</sub> = 0V One Output for MAX of One Sec.
I <sub>cc</sub>	Power Supply Current	-	155 (C)	mA	V <sub>CC</sub> = V <sub>CC</sub> MAX
		-	165 (M)	mA	

C = Commercial (-5) M = Military (-2/-8)

(1) One output at a time, otherwise 16mA.

#### A. C. SWITCHING CHARACTERISTICS (Operating)

SYMBOL	JEDEC STANDARD	OLD SYMBOL	PARAMETER	HPL-77317-5 HPL-77318-5 5V ± 5% 0°C to +75°C		HPL-77317-2/-8 HPL-77318-2/-8 5V ± 10% -55°C to +125°C		UNITS
				MIN	MAX	MIN	MAX	
TDVCH		T <sub>pd</sub>	Propagation Delay - Input to Active High Output	-	35	-	45	ns
TDVCL		T <sub>pd</sub>	Propagation Delay - Input to Active Low Output	-	35	-	45	ns

NOTE: Maximum test frequency is 5MHz with a 50% duty cycle.

### Description

The HPL-77317/318 are programmable logic devices designed to be cost effective and space saving replacements for discrete logic designs. These devices are two-level logic elements consisting of 8 product terms (AND) summed (OR) together to generate each of the 8 outputs. To achieve 8 product terms per output, the ability to force the outputs to a high impedance state has not been included. All eight outputs are permanently active.

The HPL-77317 is similar to the industry standard 16L8, and implements logic expressions of the Inverted Sum Of Products (ISOP) form.

The HPL-77318 is a similar device to the HPL-77317, but does not include the associated output inversion. It implements logic expressions of the Sum Of Products (SOP) form.

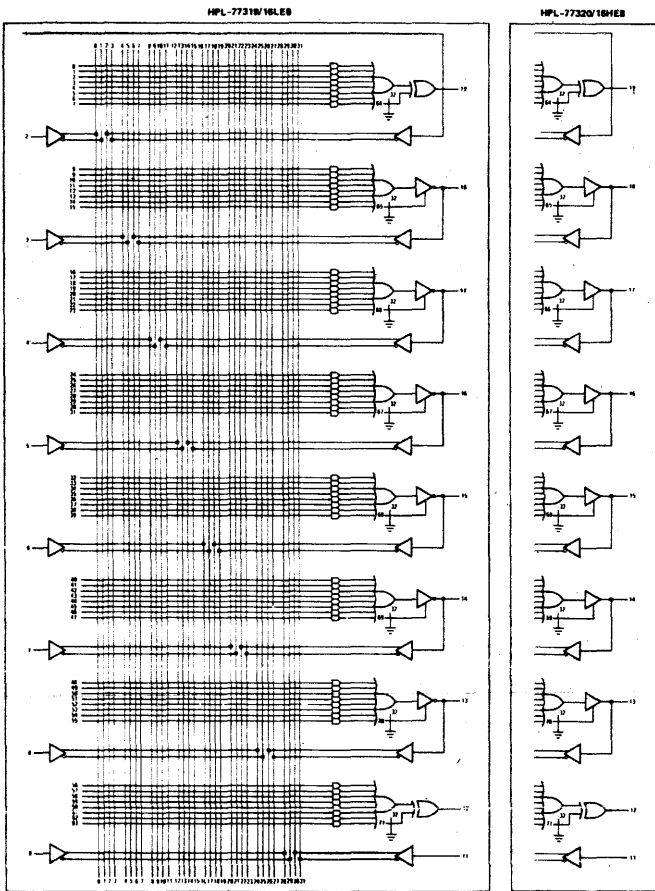
Six of the 8 outputs feature feedback into the fuse matrix, allowing output functions to be utilized for other output functions.

The HPL-77317/318 are field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

Nickel-chromium fuse technology is used on these and all other HARRIS HPL-77XXX programmable logic devices. The HPL-77317/318 are available in 20 pin slimline DIP packages with pinouts similar to the 16L8.

The HPL-77317/318 contain unique test circuitry developed by HARRIS, which is enabled at the time of manufacture to allow complete AC, DC and functional testing.

### Functional Diagram



### D.C. ELECTRICAL CHARACTERISTICS (Operating)

HPL-77319/320-5 ( $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+75^\circ C$ )  
 HPL-77319/320-2/-8 ( $V_{CC} = 5.0V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ )

SYMBOL	PARAMETER	MIN	MAX	UNITS	TEST CONDITIONS
$I_{IH}$	Input Current	-1	-	$\mu A$	$V_{IH} = V_{CC} \text{ MAX}$
$I_{IL}$	Dedicated Input	0	-25	$\mu A$	$V_{IL} = 0.4 V$
$I_{IBZ}$	Bidirectional Pin Current	-1	+40	$\mu A$	$V_{IH} = V_{CC} \text{ MAX}$
$I_{IBZ}$	Hi-Z State	0	-100	$\mu A$	$V_{CC} = V_{CC} \text{ MAX}$
$V_{IH}$	Input Threshold (1)	2.0	-	V	$V_{CC} = V_{CC} \text{ MAX}$
$V_{IL}$	Voltage	0	0.8	V	$V_{CC} = V_{CC} \text{ MIN}$
$V_{OH}$	High-Level Output Voltage (2)	2.4	-	V	$V_{CC} = V_{CC} \text{ MIN}$
$V_{OL}$	Low-Level Output Voltage (2)	-	0.5	V	$V_{IL} = 0.8V$ $V_{IH} = 2.0V$
$V_{IC}$	Input Clamp Voltage (1)	-	-1.2	V	$I_{IH} = -18mA$ , $V_{CC} = 0V$
$I_{OS}$	Output Short Circuit Current (2)	-30	-130	mA	$V_{CC} = 5.0V$ , $V_{OUT} = 0V$ One Output for MAX of One Sec.
$I_{CC}$	Power Supply Current	-	155 (C)	mA	$V_{CC} = V_{CC} \text{ MAX}$
			165 (M)	mA	

C - Commercial (-5) M - Military (-2/-8)  
 (1) These specifications apply to both Input (I) and Bidirectional (B) pins.  
 (2) These specifications apply to both Output (O) and Bidirectional (B) pins.  
 (3) One output at a time, otherwise 18mA.

### A. C. SWITCHING CHARACTERISTICS (Operating)

SYMBOL	JEDEC STANDARD	OLD SYMBOL	PARAMETER	HPL-77319-5 HPL-77320-5		HPL-77319-2/-8 HPL-77320-2/-8		UNITS
				MIN	MAX	MIN	MAX	
$T_{DQH}$		$T_{PD}$	Propagation Delay - Input or I/O to Active High Output	-	35	-	45	ns
$T_{DQL}$		$T_{PD}$	Propagation Delay - Input or I/O to Active Low Output	-	35	-	45	ns

NOTE: Maximum test frequency is 5MHz with a 50% duty cycle.

### Description

The HPL-77319/320 are programmable logic devices designed to be cost effective and space saving replacements for discrete logic designs. These devices are two-level logic elements consisting of 8 product terms (AND) summed (OR) together to generate each of the 8 outputs. An extra fuse associated with each bi-directional pin can drive it to a high impedance state allowing 6 (B0-B5) of the 8 outputs to be used as inputs.

An extra fuse associated with each output pin can be used to control output polarity which is active high on an unprogrammed part.

The HPL-77319 is similar to the industry standard 16L8, and implements logic expressions of the Inverted Sum of Products (ISOP) form.

The HPL-77320 is a similar device to the HPL-77319 but

does not include the associated output inversion. It implements logic expressions of the Sum of Products (SOP) form.

The HPL-77319/320 are field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

Nickel-chromium fuse technology is used on these and all other HARRIS HPL-77XXX programmable logic devices. The HPL-77319/320 are available in 20 pin slimline DIP packages with pinouts identical to 16L8.

The HPL-77319/320 contain unique test circuitry developed by HARRIS which is enabled at the time of manufacture to allow complete AC, DC and functional testing.



# PROGRAMMING HARRIS PROGRAMMABLE LOGIC (HPL)

Following is the programming procedure which is used for the HPL-77XXX devices. These devices are manufactured with all fuses intact. Any desired fuse can be programmed by following the simple procedure shown on the

following page. One may build a programmer to satisfy the specifications described in the table, or use any of the commercially available programmers which meet these specifications.

### PROGRAMMING SPECIFICATIONS TABLE 1

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNITS
			MIN	TYP	MAX	
VCCP	Vcc Supply During Programming <sup>1</sup>	I <sub>CCP</sub> = 550 mA min	8.25	8.5	8.75	V
ICCP	Icc Limit During Programming	V <sub>CCP</sub> = +8.50V±0.25V	550	—	1000	mA
V <sub>IH</sub>	Input Voltage (High)		2.4	—	5.5	V
V <sub>IL</sub>	Input Voltage (Low)		0.0	—	0.8	V
I <sub>IHP</sub>	Input Current (High) During Prog.	V <sub>IH</sub> = +5.5V	—	—	50	μA
I <sub>ILP</sub>	Input Current (Low) During Prog.	V <sub>IL</sub> = 0V	—	—	-500	μA
VOPF	Forced Output Voltage <sup>2</sup>	I <sub>OPF</sub> = 300mA±25mA Transient or Steady State	16.0	16.5	18.0	V
IOPF	Forced Output Current	V <sub>OPF</sub> = 17.5V±0.5V	275	300	325	mA
PW <sub>V</sub>	Verify Pulse Width		1	5	50	μS
PW <sub>P</sub>	Programming Pulse Width		90	100	110	μS
T <sub>D</sub>	Pulse Sequence Delay		1	5	50	μS
tr <sub>2</sub>	Forced Output Voltage Rise Time	10% to 90%	4	14	40	μS
tr <sub>1</sub>	Vcc Supply Rise Time	10% to 90%	2	7	20	μS
tf <sub>2</sub>	Forced Output Voltage Fall Time	90% to 10%	2	7	20	μS
tf <sub>1</sub>	Vcc Supply Fall Time	90% to 10%	1	4	10	μS
T <sub>PP</sub>	Programming Period		100	160	350	μS
FL	Fusing Attempts per Link		—	1	2	Cycle
V <sub>S</sub>	Verify Threshold <sup>3</sup>		1.4	1.5	1.6	V

#### NOTES:

1. Bypass VCC to GND with a 0.01 μF capacitor to reduce voltage spikes. The VCC power supply must be capable of handling 1 Ampere maximum.
2. Care should be taken to ensure that the voltage is maintained during the entire fusing cycle. The recommended supply is a constant current source clamped at the specified voltage limit.
3. V<sub>S</sub> is the sensing threshold of the device output voltage for a programmed link. It normally constitutes the reference voltage applied to a comparator circuit to verify a successful fusing attempt.

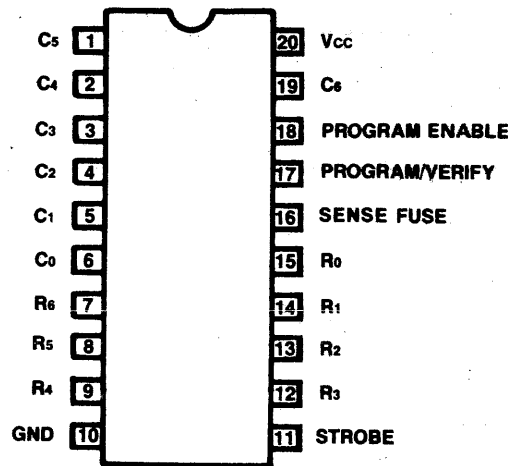


Figure 1  
EDIT MODE PINOUT  
HPL-77XXX

\*NOTE: While programming HPL-77XXX devices, no pins should be left floating. Pin 16 (SENSE FUSE) appears as an open collector output during programming. It should be tied to VCC through a 10K ohm resistor. It is recommended that all unused row and column address pins be tied to GND. Consult individual HPL data sheets for information pertaining to selection of fuses.



# HPL PROGRAMMING PROCEDURE & WAVEFORMS

## PROGRAMMING PROCEDURE

1. Set-Up:
  - a. During programming, no pins should be left floating. Pin 16 should be terminated with a 10k ohm resistor to  $V_{CC}$ .
  - b. Set GND (Pin 10) to OV.
2. Select the EDIT Mode:
  - a. Apply OV to Pin 20.
  - b. Apply  $V_{IL}$  to Pin 18.
  - c. Apply  $V_{IH}$  to Pin 11.
  - d. Apply  $V_{IL}$  to Pin 17.
3. Select the Fuse to be Programmed:
  - a. Wait  $T_D$  and select a ROW by specifying the appropriate binary ROW address according to Table 2.
  - b. Select a COLUMN by specifying the appropriate binary COLUMN address according to Table 3.
4. Program the Fuse:
  - a. Wait  $T_D$  and raise Pin 20 to  $V_{CCP}$ .
  - b. Wait  $T_D$  and raise Pin 17 to  $V_{IH}$ .
  - c. Wait  $T_D$  and raise Pin 18 to  $V_{OPF}$ .
  - d. Wait  $T_D$  and pulse Pin 11 to  $V_{IL}$  for a duration of  $PW_p$ .
  - e. Return Pin 11 to  $V_{IH}$ .
  - f. Wait  $T_D$  and lower Pin 18 to  $V_{IL}$ .
  - g. Wait  $T_D$  and lower Pin 17 to  $V_{IL}$ .
5. Verify Programmed Fuse:
  - a. Wait  $T_D$  and lower Pin 11 to  $V_{IL}$ .
  - b. At the end of  $PW_v$ , monitor Pin 16 for a level of  $V_s$ .
  - c. Return Pin 11 to  $V_{IH}$ .
  - d. Wait  $T_D$  and lower Pin 20 to OV.
  - e. If Pin 16 has indicated a  $V_{OL}$ , return to step 4 and repeat so that the fuse receives a maximum of two (2) fusing attempts.
6. Repeat Steps 3 Through 5 for All Fuses to be Blown.

## PROGRAMMING WAVEFORMS

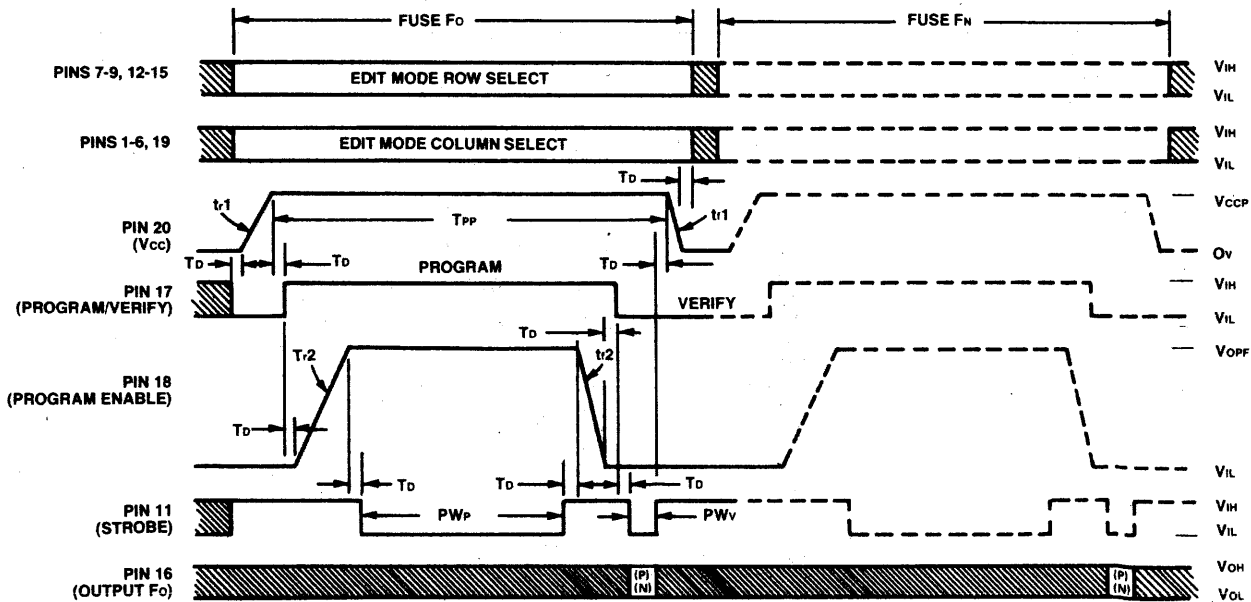


Figure 2

**Product Preview HPL™ Harris Programmable Logic**

The HPL-77061 is a proprietary device designed by Harris Semiconductor. It is a 24 pin device with 12 input pins, 6 bi-directional pins, and 4 output pins. The architecture consists of a programmable AND array, followed by a fixed OR array, followed by a programmable INVERTER. This configuration generates a device that implements logic expressions in either SUM OF PRODUCTS (SOP) or INVERTED SUM OF PRODUCTS (ISOP) form. Two fuses associated with each output and bi-directional pin allow their three state buffers to be configured as permanently active, permanently inactive, or under the control of a common CHIP SELECT (CS) signal. When under the control of the CS signal, they may be configured to be active when CS is high and inactive when low, or active when CS is low and inactive when high.

The HPL-77061 is intended to be used as an address decoder in 8, 16 or 32 bit microprocessor systems, and is known as a FIELD PROGRAMMABLE ADDRESS DECODER (FPAD). When used as an address decoder, it is possible to contiguously concatenate the address spaces of large and small system components. This allows more efficient use to be made of available memory space, as well as making the memory decoder design easier. The HPL-77061 may also be used to replace several TTL packages in many random logic designs. In both the above applications, a significant reduction in board space, power consumption, package count, and an increase in reliability may be realized. The HPL-77061 incorporates Harris designed test circuitry which allows complete parametric and logical testing of unprogrammed devices at the time of manufacture. The benefits for the user of this exhaustive testing are very high POST PROGRAMMING FUNCTIONAL YIELD (95% guaranteed; 97% typically) and pretested AC and DC parametrics.

The HPL-77061 uses the standard HPL programming algorithm which is also used for all Harris Programmable Logic parts. Special test fuses are included in the HPL-77061. These are selectively programmed during post manufacturing testing. This allows all the programming circuitry to be tested for correct operation.

The Harris policy is to ship pretested devices, removing the test burden from the small user and assuring the large user the highest possible quality.

- PROPAGATION DELAY 25ns MAX COMMERCIAL
- SUPPLY CURRENT 140mA MAX COMMERCIAL
- PACKAGE 24 PIN SLIMLINE  
(POWER PLASTIC & CERAMIC)
- TEMPERATURE OPTIONS COMMERCIAL, MILITARY & 883B
- AC/DC SPECS PRETESTED
- FUNCTIONALITY PRETESTED
- PROGRAMMING GENERIC
- TTL COMPATIBLE



### CMOS/NMOS RAMs - Cross Reference

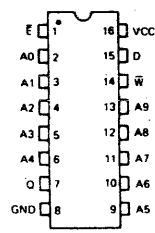
Description	HARRIS	AMI	Fujitsu	Hitachi	IDT	Intersil	Micro Power Systems	Mitsubishi
1K CMOS RAMs	HM-6508 (1K × 1) 16-pin synch	6508	8401			6508	6508	
	HM-6518 (1K × 1) 18-pin synch	6518				6518	6518	
	HM-6551 (256 × 4) 22-pin synch					6551		
	HM-6561 (256 × 4) 18-pin synch					6561		
4K CMOS RAMs	HM-6504 (4K × 1) 18-pin synch	6504	8404	4315 6147		6504	6504	
	HM-6514 (1K × 4) 18-pin synch	6514	8414	4334 6148		6514	6514	58981
16K CMOS RAMs	HM-6516 (2K × 8) 24-pin synch	6516						
	HM-65162 (2K × 8) 24-pin asynch		8416	6116	6116			5117
	HM-65172 (2K × 8) 24-pin asynch		8418	6177				5116

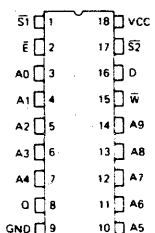
Description	HARRIS	Motorola	National	NEC	OKI	RCA	Toshiba	NMOS Memories (Generic)
1K CMOS RAMs	HM-6508 (1K × 1) 16-pin synch	6508	6508 74C929	443		6508 1821	5508	2125 4015
	HM-6518 (1K × 1) 18-pin synch	6518	6518 74C930					
	HM-6551 (256 × 4) 22-pin synch		6551 74C920			1822	5101	2101
	HM-6561 (256 × 4) 18-pin synch							2111
4K CMOS RAMs	HM-6504 (4K × 1) 18-pin synch	6504	6504		5104		5504	2141/47 315D 4104 4404
	HM-6514 (1K × 4) 18-pin synch	6514	6514	444	5114 5115	5114	5114	2114 2148/49 4045 314A
16K CMOS RAMs	HM-6516 (2K × 8) 24-pin synch		6516					
	HM-65162 (2K × 1) 24-pin asynch	65116	6116	446	5128	6116	5517	2116 2016 4016 4802 2128 8128 58725
	HM-65172 (2K × 8) 24-pin asynch			449			5516	

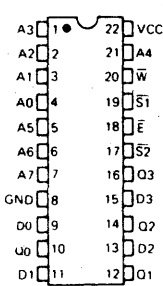
Harris Semiconductor

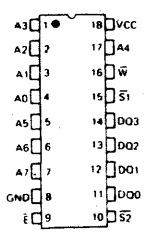
MEMORY

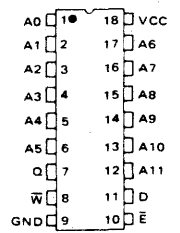


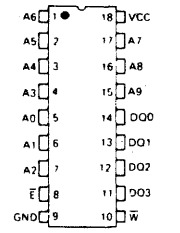
1024 × 1 — 1K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-6508B	180 ns	4 mA/MHz	10 $\mu$ A	National 74C929 Intersil 6508  AMI 56508	
HM-6508	250 ns	4 mA/MHz	10 $\mu$ A		

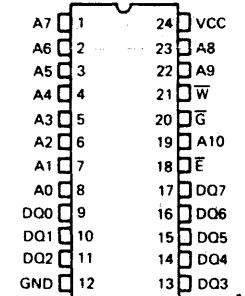
1024 × 1 — 1K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-6518B	180 ns	4 mA/MHz	10 $\mu$ A	National 74C930 Intersil 6518	
HM-6518	250 ns	4 mA/MHz	10 $\mu$ A		

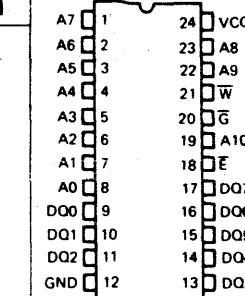
256 × 4 — 1K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-6551B	220 ns	4 mA/MHz	10 $\mu$ A	Intersil 6551	
HM-6551	300 ns	4 mA/MHz	10 $\mu$ A		

256 × 4 — 1K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-6561B	220 ns	4 mA/MHz	10 $\mu$ A	Intersil 6561	
HM-6561	300 ns	4 mA/MHz	10 $\mu$ A		

4096 × 1 — 4K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-6504S	120 ns	7 mA/MHz	25 $\mu$ A	Fujitsu 8404 Oki 5104 National 6504	
HM-6504B	200 ns	7 mA/MHz	25 $\mu$ A		
HM-6504	300 ns	7 mA/MHz	25 $\mu$ A		

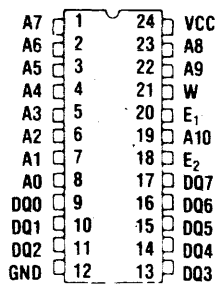
1024 × 4 — 4K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-6514S	120 ns	7 mA/MHz	25 $\mu$ A	Fujitsu 6514 Hitachi 4334 NEC 444 RCA 5114 Toshiba 5514 National 6514	
HM-6514B	200 ns	7 mA/MHz	25 $\mu$ A		
HM-6514	300 ns	7 mA/MHz	25 $\mu$ A		

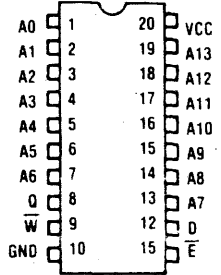
2048 × 8 — 16K Synchronous					Pinout
Part Number	Access Time	Power Supply Current		Similar To	
		Operating	Standby		
HM-6516B	120 ns	10 mA/MHz	50 $\mu$ A	Hitachi 6116 Toshiba 5517	
HM-6516	200 ns	10 mA/MHz	50 $\mu$ A		

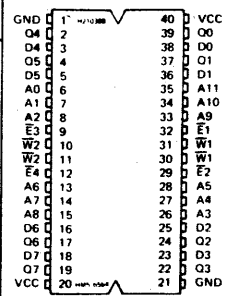
2048 × 8 — 16K Asynchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-65162S	55 ns	70 mA	100 $\mu$ A	Fujitsu 8416 Hitachi 6116 NEC 446 Toshiba 5517 National 6116 IDT 6116	
HM-65162B	70 ns	70 mA	50 $\mu$ A		
HM-65162	90 ns	70 mA	100 $\mu$ A		
HM-65162C	90 ns (Ind) 120 ns (Mil)	70 mA	1 mA		

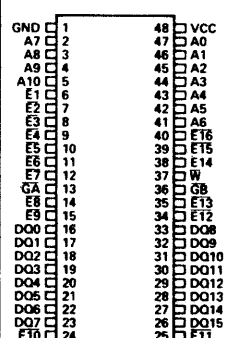
Harris Semiconductor

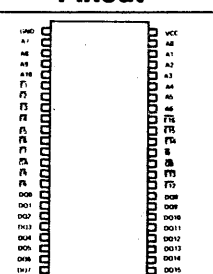
MEMORY

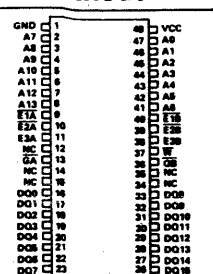
2048 x 8 — 16K Asynchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
HM-65172S	55 ns	70 mA	100 $\mu$ A	Fujitsu 8418 Hitachi 6117 Toshiba 5516 NEC 449	
HM-65172B	70 ns	70 mA	50 $\mu$ A		
HM-65172	90 ns	70 mA	100 $\mu$ A		
HM-65172C	90 ns (Ind) 120 ns (Mil)	70 mA	1 mA		

16384 x 1 — 16K Asynchronous					Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin	
		Operating	Standby		
<b>COMING SOON</b> HM-65262B	55 ns	60 mA	50 $\mu$ A	Hitachi 6167 IDT 6167	
HM-65262	70 ns	60 mA	50 $\mu$ A		

LCC RAM Module — 16384 × 4/8192 × — 64K Synchronous				Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin
		Operating	Standby	
HM5-6564	350 ns	56/28 mA	800 μA	
HM5-6564-5	450 ns	60/30 mA	4.0 μA	

LCC RAM Module - 16384 × 16/32768 × 8 — 256K Synchronous				Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin
		Operating	Standby	
HM-92560	150 ns	30/15 mA	50 μA	
HM-92560-5	250 ns	35/20 mA	3.5 mA	

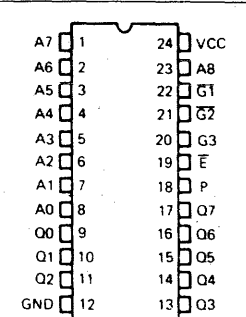
LCC Unbuffered RAM Module - 16384 × 16/32768 × 8 — 256K Asynchronous				Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin
		Operating	Standby	
<b>COMING SOON</b> HM-92562	90 ns	75 mA	500 μA	

LCC Buffered RAM Module - 16384 × 16/32768 × 8 — 256K Synchronous				Pinout
Part Number	Access Time	Power Supply Current		Replaces Pin for Pin
		Operating	Standby	
HM-92570	250 ns 300 ns	30/15 mA 35/20 mA 35/20 mA	600 μA 3.5 mA	

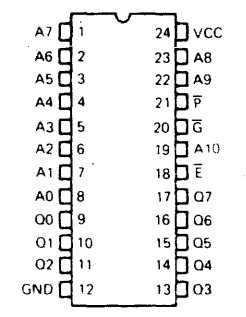
Harris Semiconductor

**MEMORY**

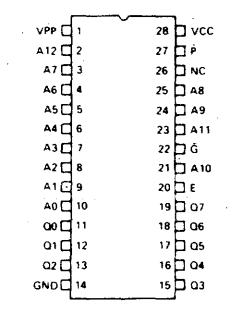
512 × 8 — 4K Synchronous					Pinout
Part Number	Memory Element	Access Time	Power Supply Current		Replaces Pin for Pin
			Operating	Standby	
HM-6641	Fuse	250 ns	10 mA/MHz	100 μA	Bipolar 7641 Signetics 82S141



2048 × 8 — 16K Synchronous					Pinout
Part Number	Memory Element	Access Time	Power Supply Current		Replaces Pin for Pin
			Operating	Standby	
HM-6616B HM-6616	Fuse Fuse	90 ns 120 ns	15 mA/MHz 15 mA/MHz	100 μA 100 μA	NMOS 2716 National 27C16 National 6716 Intersil 6716



8192 × 8 — 64K Asynchronous					Pinout
Part Number	Memory Element	Access Time	Power Supply Current		Replaces Pin for Pin
			Operating	Standby	
<b>COMING SOON</b> HM-6664	Fuse	175 ns	10 mA/MHz	100 μA	NMOS 2764 Fujitsu 27C64



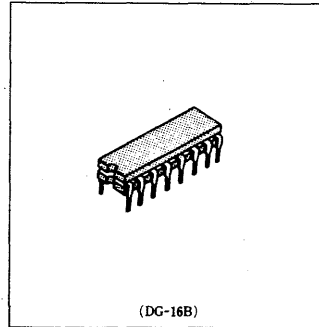
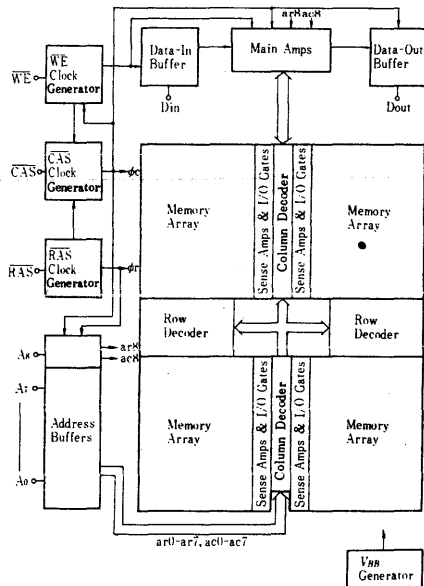
# HM50256-12, HM50256-15 — Preliminary HM50256-20

262144-word × 1-bit Dynamic Random Access Memory

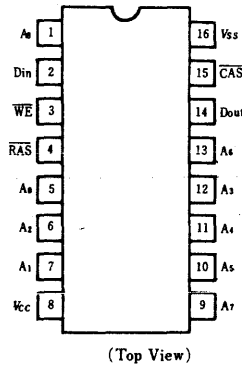
## FEATURES

- Industry Standard 16-Pin DIP
- Single 5V ( $\pm 10\%$ )
- On chip substrate bias generator
- Low Power: 350mW active, 23mW standby
- High speed: Access Time 120ns/150ns/200ns(max.)
- Common I/O capability using early write operation
- Page mode capability
- TTL compatible
- 256 refresh cycles.....(4 ms)
- 3 variation of refresh ... RAS only refresh  
CAS before RAS refresh  
Hidden refresh

## BLOCK DIAGRAM



## PIN ARRANGEMENT



A <sub>0</sub> ~A <sub>4</sub>	Address Inputs
CAS	Column Address Strobe
Din	Data In
Dout	Data Out
RAS	Row Address Strobe
WE	Read/Write Input
V <sub>cc</sub>	Power (+5V)
V <sub>ss</sub>	Ground
A <sub>0</sub> ~A <sub>7</sub>	Refresh Address Inputs

## ABSOLUTE MAXIMUM RATINGS

- Voltage on any pin relative to V<sub>SS</sub> ..... -1V to +7V  
 Operating temperature, T<sub>a</sub> (Ambient) ..... 0°C to +70°C  
 Storage temperature ..... -65°C to +150°C  
 Power dissipation ..... 1W  
 Short circuit output current ..... 50mA

## RECOMMENDED DC OPERATING CONDITIONS (T<sub>a</sub>=0 to +70°C)

Parameter	Symbol	min	typ	max	Unit	Note
Supply Voltage	V <sub>CC</sub>	4.5	5.0	5.5	V	1
Input High Voltage	V <sub>IH</sub>	2.4	—	6.5	V	1
Input Low Voltage	V <sub>IL</sub>	-1.0	—	0.8	V	1

Note) 1. All voltages referenced to V<sub>SS</sub>

Note) The specifications of this device are subject to change without notice. Please contact your nearest Hitachi Sales Dept. regarding specifications.

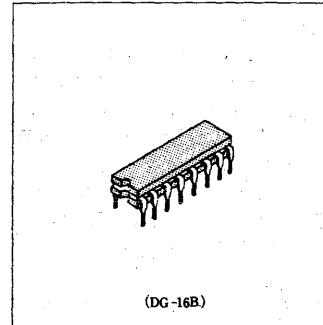


# HM50257-12, HM50257-15, — Preliminary HM50257-20

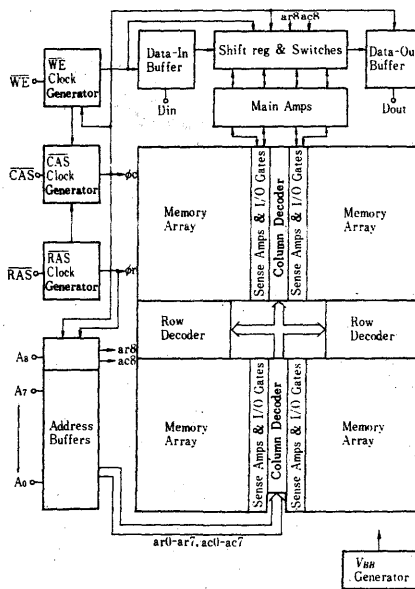
262144-word × 1-bit Dynamic Random Access Memory

## ■ FEATURES

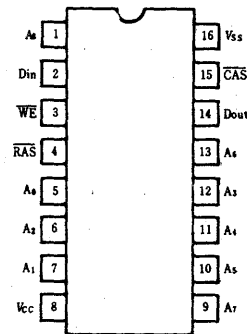
- Industry standard 16-pin DIP
- Single 5V (±10%)
- On chip substrate bias generator
- Low Power: 350mW active, 20mW standby
- High speed: Access Time 120ns/150ns/200ns (max.)
- Common I/O capability using early write operation
- Nibble mode capability
- Indefinite Dout hold using  $\overline{\text{CAS}}$  control
- TTL compatible
- 256 refresh cycles . . . . . (4ms)
- 3 Variations of refresh:  $\overline{\text{RAS}}$  only refresh,  $\overline{\text{CAS}}$  before  $\overline{\text{RAS}}$  refresh, Hidden refresh



## ■ BLOCK DIAGRAM



## ■ PIN ARRANGEMENT



(Top View)

A <sub>0</sub> ~A <sub>7</sub>	Address Inputs
CAS	Column Address Strobe
Din	Data In
Dout	Data Out
RAS	Row Address Strobe
WE	Read/Write Input
V <sub>cc</sub>	Power (+5V)
V <sub>ss</sub>	Ground
A <sub>0</sub> ~A <sub>7</sub>	Refresh Address Inputs

## ■ ABSOLUTE MAXIMUM RATINGS

- Voltage on any pin relative to V<sub>SS</sub> . . . . . -1V to +7V
- Operating temperature, T<sub>a</sub> (Ambient) . . . . . 0°C to +70°C
- Storage temperature . . . . . -65°C to +150°C
- Power dissipation . . . . . 1W
- Short circuit output current . . . . . 50mA

## ■ RECOMMENDED DC OPERATING CONDITIONS (T<sub>a</sub>=0 to +70°C)

Parameter	Symbol	min	typ	max	Unit	Note
Supply Voltage	V <sub>cc</sub>	4.5	5.0	5.5	V	1
Input High Voltage	V <sub>IH</sub>	2.4	—	6.5	V	1
Input Low Voltage	V <sub>IL</sub>	-1.0	—	0.8	V	1

Note 1) All voltages referenced to V<sub>SS</sub>.

Note)

The specifications of this device are subject to change without notice. Please contact your nearest Hitachi Sales Dept. regarding specifications.



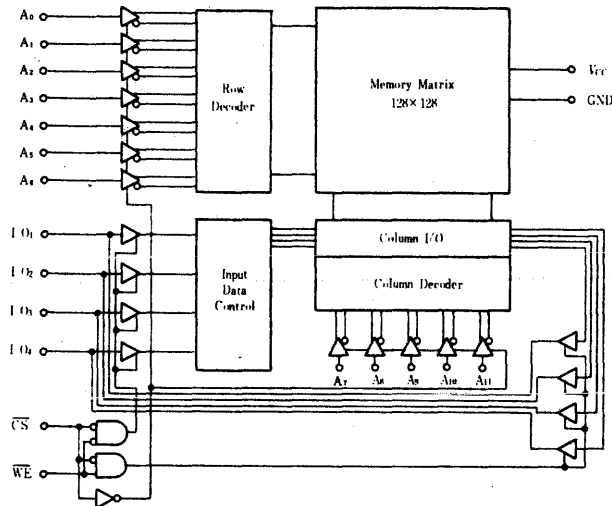
# HM6168H-45, HM6168H-55, — Preliminary HM6168H-70, HM6168HP-45, HM6168HP-55, HM6168HP-70

4096-word × 4-bit High Speed Static CMOS RAM

## ■ FEATURES

- High Speed: Fast Access Time 45/55/70 ns (max.)
- Single +5V Supply and High Density 20 Pin Package
- Low Power Standby and Low Power Operation;  
100μW typ. (Standby), 250mW typ. (Operation)
- Completely Static Memory  
No Clock or Timing Strobe Required
- Equal Access and Cycle Times
- Directly TTL Compatible — All Inputs and Outputs

## ■ FUNCTIONAL BLOCK DIAGRAM

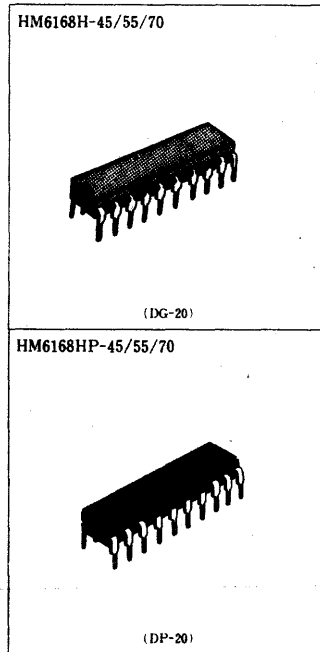


## ■ ABSOLUTE MAXIMUM RATINGS

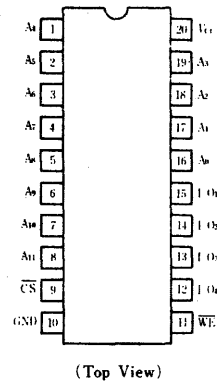
Item	Symbol	Rating	Unit
Voltage on Any Pin Relative to GND	$V_{IN}$	-3.5 to +7.0	V
Power Dissipation	$P_T$	1.0	W
Operating Temperature	$T_{op}$	0 to +70	°C
Storage Temperature (Ceramic)	$T_{stc}$	-65 to +150	°C
Storage Temperature (Plastic)	$T_{stp}$	-55 to +125	°C
Temperature under Bias	$T_{bss}$	-10 to +85	°C

\* Pulse Width 20ns, DC = -0.5V

Note) The specifications of this device are subject to change without notice.  
Please contact your nearest Hitachi Sales Dept. regarding specifications.



## ■ PIN ARRANGEMENT



(Top View)



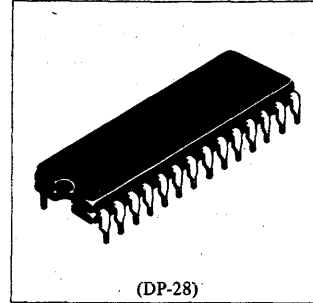


# HM6264P-10, HM6264P-12, HM6264P-15

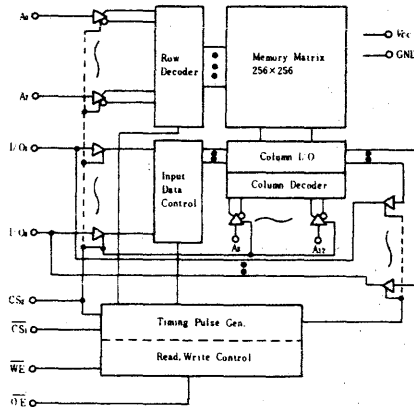
8192-word x 8-bit High Speed Static CMOS RAM

## ■ FEATURES

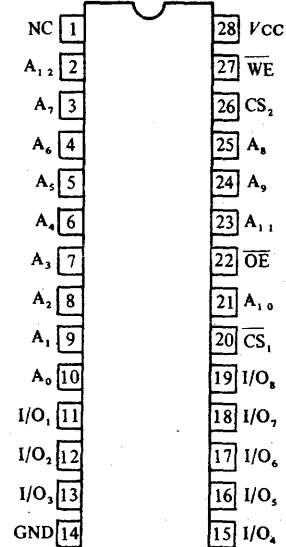
- Fast access Time 100ns/120ns/150ns (max.)
- Low Power Standby Standby: 0.1mW (typ.)
- Low Power Operation Operating: 200mW (typ.)
- Single +5V Supply
- Completely Static Memory . . . . . No clock or Timing Strobe Required
- Equal Access and Cycle Time
- Common Data Input and Output, Three State Output
- Directly TTL Compatible: All Input and Output
- Standard 28pin Package Configuration
- Pin Out Compatible with 64K EPROM HN482764



## ■ BLOCK DIAGRAM



## ■ PIN ARRANGEMENT



(Top View)

## ■ ABSOLUTE MAXIMUM RATINGS

Item	Symbol	Rating	Unit
Terminal Voltage *	$V_T$	-0.5 ** to +7.0	V
Power Dissipation	$P_T$	1.0	W
Operating Temperature	$T_{opr}$	0 to +70	°C
Storage Temperature	$T_{stg}$	-55 to +125	°C
Storage Temperature (Under Bias)	$T_{bias}$	-10 to +85	°C

\* With respect to GND. \*\* Pulse width 50ns: -3.0V

## ■ TRUTH TABLE

$\overline{WE}$	$CS_1$	$CS_2$	$\overline{OE}$	Mode	I/O Pin	$V_{CC}$ Current	Note
X	H	X	X	Not Selected (Power Down)	High Z	$ISB, ISB1$	
X	X	L	X		High Z	$ISB, ISB2$	
H	L	H	H	Output Disabled	High Z	$ICC, ICC1$	
H	L	H	L	Read	Dout	$ICC, ICC1$	
L	L	H	H	Write	Din	$ICC, ICC1$	Write Cycle (1)
L	L	H	L		Din	$ICC, ICC1$	Write Cycle (2)

X : Don't care.



# HN27C64G-20 HN27C64G-25, HN27C64G-30

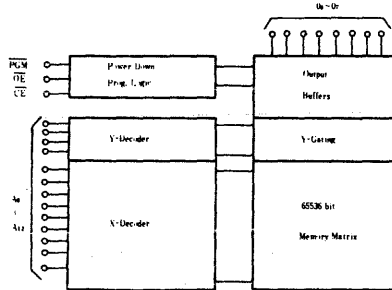
## 8192-word X 8-bit UV Erasable and Programmable Read Only Memory

The HN27C64G is a 8192 word by 8-bit erasable and electrically programmable ROM. This device is packaged in a 28 pin dual-in-line package with transparent lid. The transparent lid on the package allows the memory content to be erased with ultraviolet light.

### ■ FEATURES

- Single Power Supply ..... +5V  $\pm$  5%
- Simple Programming..... Program Voltage: +21V D.C.  
Program with one 50ms Pulse
- Static..... No clocks Required
- Inputs and Outputs TTL Compatible During Both Read and Program Mode.
- Access Time..... HN27C64G-20 200ns Max.  
HN27C64G-25 250ns Max.  
HN27C64G-30 300ns Max.
- High Performance Programming Available
- Low Power Dissipation..... 40mW/MHz (Active Mode)  
525 $\mu$ W Max. (Stand-by Mode)

### ■ BLOCK DIAGRAM



### ■ MODE SELECTION

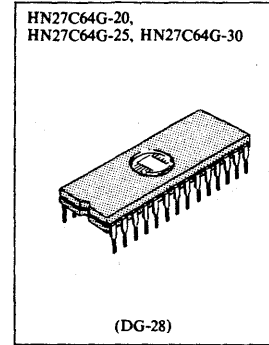
Mode	Pins	CE (20)	OE (22)	PGM (27)	V <sub>PP</sub> (1)	V <sub>CC</sub> (28)	Outputs (11~13, 15~19)
Read		V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>CC</sub>	V <sub>CC</sub>	Dout
Stand-by		V <sub>IH</sub>	X	X	V <sub>CC</sub>	V <sub>CC</sub>	High Z
Program		V <sub>IL</sub>	X	V <sub>IL</sub>	V <sub>PP</sub>	V <sub>CC</sub>	Din
Program Verify		V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>PP</sub>	V <sub>CC</sub>	Dout
Program Inhibit		V <sub>IH</sub>	X	X	V <sub>PP</sub>	V <sub>CC</sub>	High Z

X : don't care

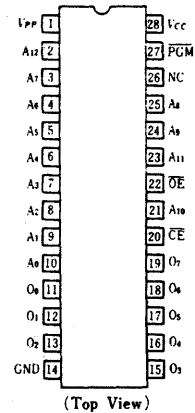
### ■ ABSOLUTE MAXIMUM RATINGS

Item	Symbol	Value	Unit
Operating Temperature Range	T <sub>OPR</sub>	0 to +70	°C
Storage Temperature Range	T <sub>STG</sub>	-65 to +125	°C
All Input and Output Voltage*	V <sub>IN</sub> , V <sub>OUT</sub>	-0.6 to +7.0	V
V <sub>PP</sub> Voltage	V <sub>PP</sub>	-0.6 to +25	V
V <sub>CC</sub> Voltage*	V <sub>CC</sub>	-0.6 to +7.0	V

\* : with respect to GND



### ■ PIN ARRANGEMENT



Hitachi America  
MEMORY



# HN4827128G-25, HN4827128G-30, HN4827128G-45

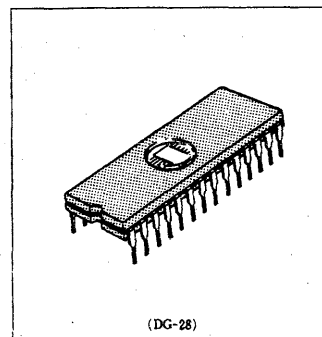
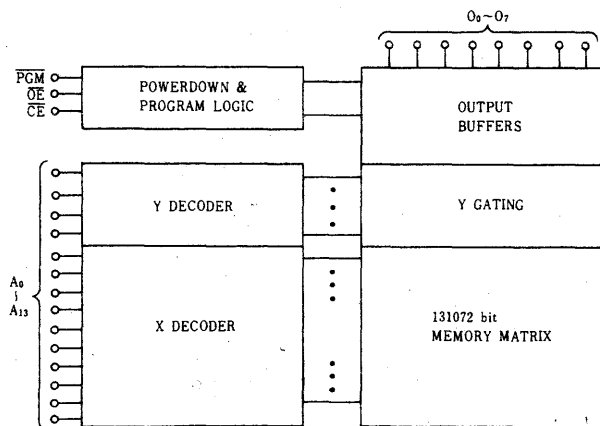
## 16384-Word x 8-bit UV Erasable and Programmable Read Only Memory

The HN4827128 is a 16384 word by 8 bit erasable and electrically programmable ROM. This device is packaged in a dual-in-line package with transparent lid. The transparent lid allows the user to expose the chip to ultraviolet light to erase the bit pattern, whereby a new pattern can then be written into the device.

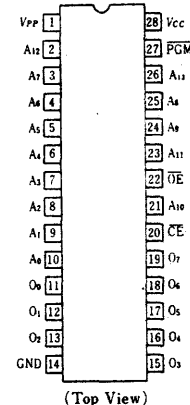
### FEATURES

- Single Power Supply ..... +5V ± 5%
- Simple Programming ..... Program Voltage: +21V DC  
Program with One 50ms Pulse
- Static ..... No Clocks Required  
Inputs and Outputs TTL Compatible During Both Read and Program Mode.
- Access Time ..... 250ns/300ns/450ns
- Absolute Max. Rating of Vpp Pin ..... 26.5V
- Low Stand-by Current ..... 35mA
- High Performance Programming Available
- Compatible with INTEL 27128

### BLOCK DIAGRAM



### PIN ARRANGEMENT



### MODE SELECTION

MODE	Pins	$\overline{CE}$ (20)	$\overline{OE}$ (22)	PGM (27)	V <sub>PP</sub> (1)	V <sub>CC</sub> (28)	Outputs (11~13, 15~19)
Read		V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>CC</sub>	V <sub>CC</sub>	Dout
Stand by		V <sub>IH</sub>	X	X	V <sub>CC</sub>	V <sub>CC</sub>	High Z
Program		V <sub>IL</sub>	X	V <sub>IL</sub>	V <sub>PP</sub>	V <sub>CC</sub>	Din
Program Verify		V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>PP</sub>	V <sub>CC</sub>	Dout
Program Inhibit		V <sub>IH</sub>	X	X	V <sub>PP</sub>	V <sub>CC</sub>	High Z

Note) The specifications of this device are subject to change without notice.  
Please contact your nearest Hitachi's Sales Dept. regarding specifications.

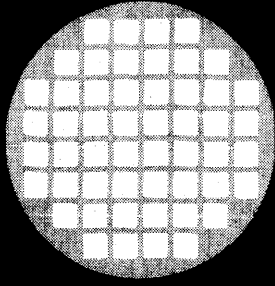




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# High Performance 16Kx1 Static RAM

## IMS1400

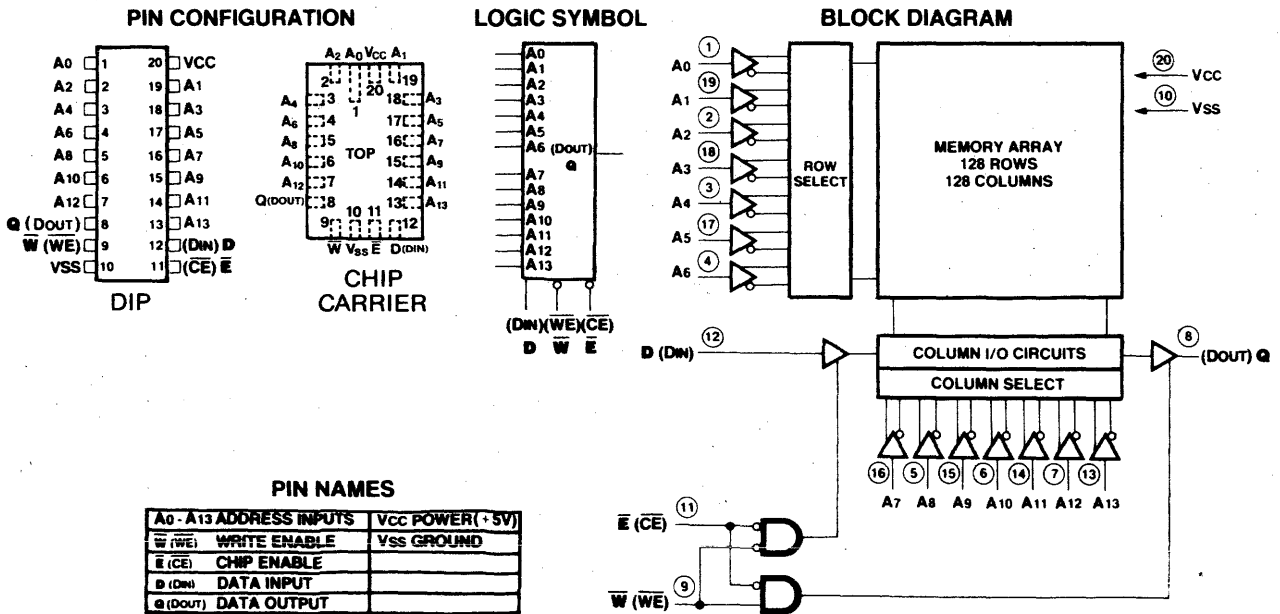
## IMS1400L

### FEATURES

- 35, 45 and 55ns Chip Enable access
- Maximum active power 660mW
- Maximum standby power 110mW
- Single 5 volt  $\pm$  10% supply
- $\bar{E}$  ( $\bar{CE}$ ) power down function
- TTL compatible inputs and output
- Fully static - no clocks for timing
- Three-state output

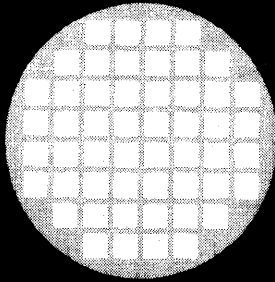
### FEATURES

- 70 and 100ns Chip Enable access
- Maximum active power 495mW
- Maximum standby power 83mW
- Single 5V  $\pm$  10% Supply
- $\bar{E}$  power down function
- TTL compatible inputs and output
- Fully static - no clocks for timing
- Three-state output



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# High Performance 4Kx4 Static RAM

**IMS1420**  
**IMS1421**

**IMS1420L**

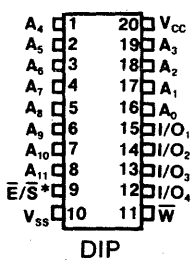
**FEATURES**

- 4K x 4 Bit Organization
  - 40nsec and 50nsec Address Access Times
  - 605mW Maximum Power Dissipation
  - Fully TTL Compatible
  - Common Data Inputs & Outputs
  - 20-pin, 300-mil DIP
  - Single +5V ± 10% Operation
- IMS1420** {
- 45nsec and 55nsec Chip Enable Access Times
  - Power Down Function
  - 165mW Maximum Standby Power
- IMS1421** {
- High Speed Chip Select Function
  - 30nsec and 40nsec Chip Select Access Times

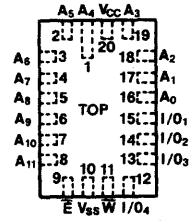
**FEATURES**

- 4K x 4 Bit Organization
- 70 and 100ns Chip Enable Access Times
- 65 and 95ns Address Access Times
- 495mW Maximum Power Dissipation
- 83mW Maximum Standby Power
- Fully TTL Compatible
- Common Data Inputs & Outputs
- 20-pin, 300-mil Plastic DIP
- Single +5V ± 10% Operation
- Power Down Function

**PIN CONFIGURATION**

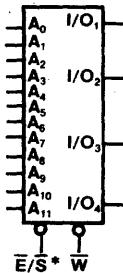


DIP

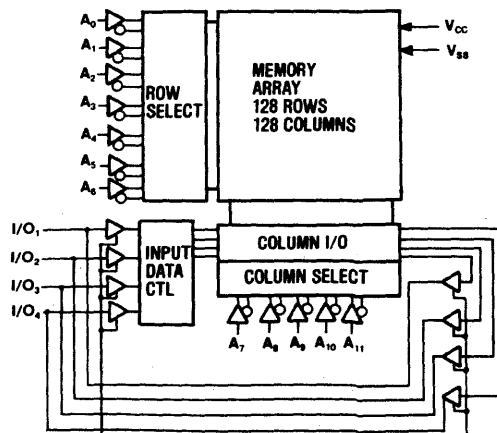


CHIP CARRIER

**LOGIC SYMBOL**



**BLOCK DIAGRAM**



**PIN NAMES**

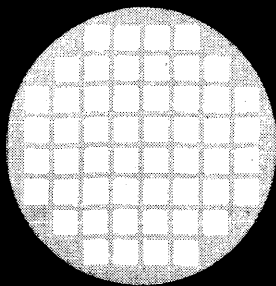
A <sub>0</sub> -A <sub>11</sub>	ADDRESS INPUTS	V <sub>CC</sub>	POWER (+5V)
W	WRITE ENABLE	V <sub>SS</sub>	GROUND
E*	CHIP ENABLE		
S*	CHIP SELECT		
I/O	DATA IN/OUT		

\* E IMS1420 ONLY  
S IMS1421 ONLY

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Inmos  
MEMORY



**inmos**

# High Performance 64K Dynamic RAM

**IMS2600 64Kx1**

**IMS2620 16Kx4**

\*No Page Mode Operation

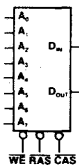
**FEATURES**

- High Speed,  $\overline{\text{RAS}}$  Access of 100, 120 and 150ns
- Cycle Times of 160, 190 and 230ns
- Low Power:  
22mW Standby  
303mW Active (350ns Cycle Time)  
413mW Active (190ns Cycle Time)
- Single +5V  $\pm$  10% Power Supply
- On-Chip refresh using  $\overline{\text{CAS}}$ -before- $\overline{\text{RAS}}$ , Pin 1 left as N/C for 256K expansion
- Indefinite  $\text{D}_{\text{OUT}}$  Hold Under  $\overline{\text{CAS}}$  Control
- Industry Standard 16 Pin Configuration
- Nibble-Mode Capability (High Speed 4 Bit Serial Mode)
- 4ms/256 Cycle Refresh
- All Inputs and Output TTL Compatible
- Read, Write, Read-Modify-Write Capability both on Single Bit and in Nibble Mode Operation
- $\overline{\text{RAS}}$ -Only Refresh Capability
- Common I/O Capability using "Early-Write"

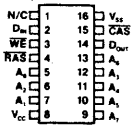
**Pin Names**

$A_0 - A_7$	ADDRESS INPUTS
$\overline{\text{CAS}}$	COLUMN ADDRESS STROBE
$\overline{\text{RAS}}$	ROW ADDRESS STROBE
$\text{D}_{\text{IN}}$	DATA IN
$\text{D}_{\text{OUT}}$	DATA OUT
$\overline{\text{WE}}$	WRITE ENABLE
$\text{V}_{\text{CC}}$	+5 VOLT SUPPLY INPUT
$\text{V}_{\text{SS}}$	GROUND

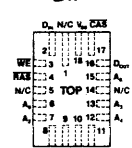
**Logic Symbol**



**Pin Configuration**

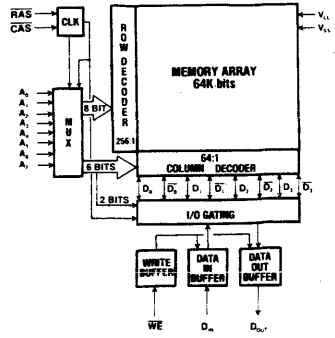


DIP

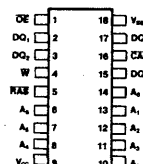


Chip Carrier

**Block Diagram**



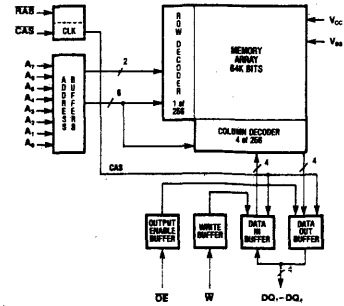
**Pin Configuration**



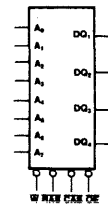
**Pin Names**

$A_0 - A_7$	ADDRESS INPUTS
$\overline{\text{CAS}}$	COLUMN ADDRESS STROBE
$\overline{\text{RAS}}$	ROW ADDRESS STROBE
$\text{DQ}_0 - \text{DQ}_3$	DATA IN/DATA OUT
$\overline{\text{OE}}$	OUTPUT ENABLE
$\overline{\text{W}}$	WRITE ENABLE
$\text{V}_{\text{CC}}$	+5V SUPPLY
$\text{V}_{\text{SS}}$	GROUND

**Block Diagram**

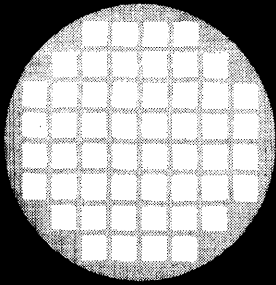


**Logic Symbol**



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# IMS2630

## High Performance 8Kx8 Dynamic RAM

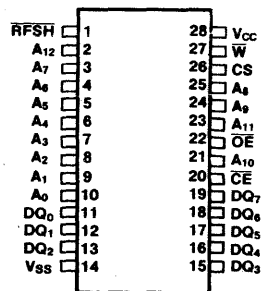
### FEATURES

- 8Kx8 Byte-Wide Organization
- High Speed Chip Enable Access of 120, 150 and 200ns
- 28-Pin Package—JEDEC Standard Pinout
- 256 Cycle/4ms Refresh
- On-Chip Refresh Counter
- Hidden Refresh Using Pin 1 Refresh
- Single +5V ± 10% Supply
- Latched Chip Select and Address
- High Speed Output Enable Control
- Low Power
  - 42mW Standby
  - 358mW Max @ 190ns cycle
  - 275mW Max @ 240ns cycle
  - 220mW Max @ 310ns cycle

### TYPICAL APPLICATIONS

- Cost-Effective Alternative to 8Kx8 SRAMs
- Easily Interfaced with Microprocessors
- Buffer Memory
- Controllers
- Terminals
- Printers
- Displays
- Personal Computers
- Games
- Instrumentation

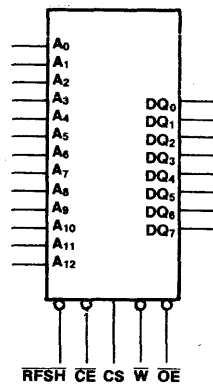
### PIN CONFIGURATION



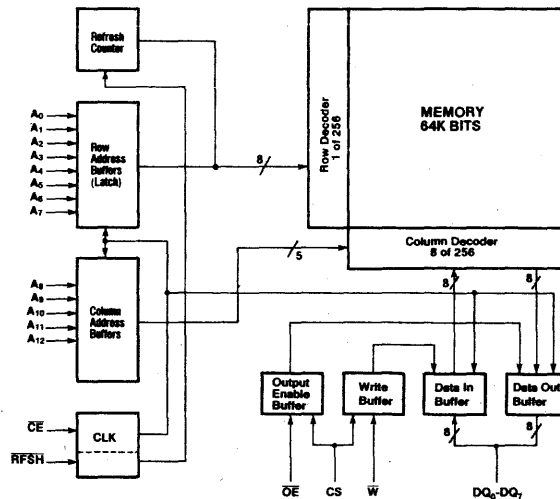
### PIN NAMES

A <sub>0</sub> -A <sub>12</sub>	ADDRESS INPUTS	DQ <sub>0</sub> -DQ <sub>7</sub>	DATA IN/DATA OUT
CE	CHIP ENABLE	V <sub>CC</sub>	+5V SUPPLY
W	WRITE ENABLE	V <sub>SS</sub>	GROUND
OE	OUTPUT ENABLE	RFSH	REFRESH ENABLE

### LOGIC SYMBOL



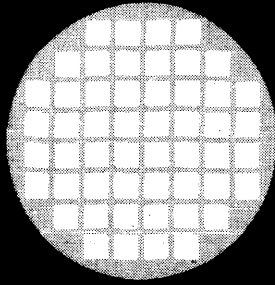
### BLOCK DIAGRAM



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# IMS3630

## 8Kx8 EEPROM

### FEATURES

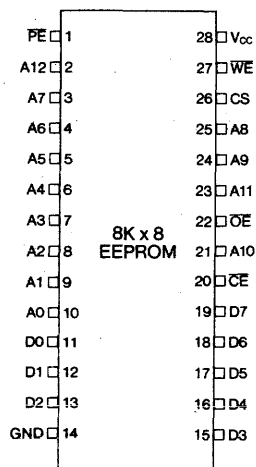
- High Speed, Chip Enable Access of 200ns
- +5 Volt Only Operation For All Modes
- Advanced Nitrox Process
- Row Or Full Array Erase Mode
- Single Byte Or Multiple Byte Modification
- Row Programming—Provides Up To 64 Times Faster Programming
- Latched Addresses and Data
- Latched Program/Erase Allows Processor To Perform Other Functions During Non-Volatile Modification
- Clocked Operation Provides Low Standby Power
- Industry Standard 28 Pin Dual-In-Line Pinout

### TYPICAL APPLICATIONS

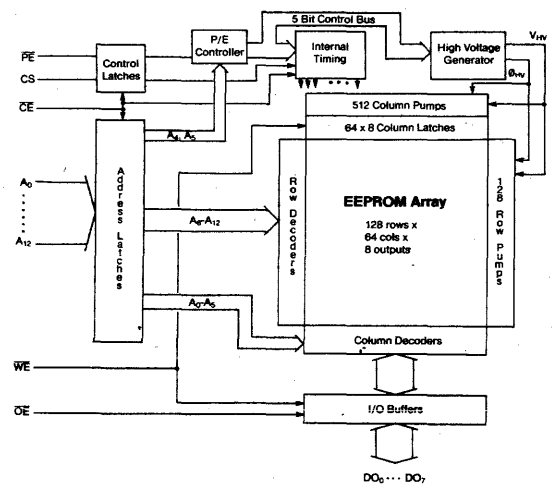
- Computer Systems
- Point of Sale Terminals
- Communication Systems
- Process Controllers
- Telemetry Systems
- Control Systems
- Instrumentation

The IMS3630 has been optimized for in-system operation. All modes of operation, including Chip Erase, are executed with a single +5 volt supply and standard TTL signal levels. Furthermore, all programming and erasing operations are controlled by using a simple sequence of initiate and terminate cycles. This eliminates the need to hold control signals, addresses, or data during the extended intervals required for programming or erasing.

### PIN CONFIGURATION



### FUNCTIONAL BLOCK DIAGRAM



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Integrated Device Technology, Inc.

# CMOS STATIC RAMS 16K (2K x 8 BIT)

# IDT6116S IDT6116L

MILITARY / INDUSTRIAL / COMMERCIAL TEMPERATURE RANGES

### FEATURES:

- High-speed (equal access and cycle time)
  - MILITARY/INDUSTRIAL
    - IDT6116S 90/120/150 ns (max.)
    - IDT6116L 90/120/150 ns (max.)
  - COMMERCIAL
    - IDT6116S 70/90/120 ns (max.)
    - IDT6116L 90/120/150 ns (max.)
- Low power consumption
 

IDT6116S	IDT6116L
Active: 180 mW (Typ.)	Active: 160 mW (Typ.)
Standby: 100 μW (Typ.)	Standby: 20 μW (Typ.)
- Battery backup operation — 2V data retention voltage
- Produced with advanced CEMOSTM I high-performance technology
- CEMOSTM I process virtually eliminates alpha particle soft-error rates (with no organic die coatings)
- Single 5V (±10%) power supply
- Input and output directly TTL compatible
- Three-state output
- Static operation: no clocks or refresh required
- Standard 24-pin dual-in-line ceramic sidebraze package or 28-pin and 32-pin LCC
- Pin compatible with standard 16K static RAM and EPROM
- Standard product 100% screened to MIL-STD-883, Class C
- Military product available 100% screened to Class B

### DESCRIPTION:

The IDT6116 is a 16,384-bit high-speed static RAM organized as 2K x 8. It is fabricated using IDT's high-performance, high-reliability technology—CEMOSTM I. This state-of-the-art technology, combined with innovative circuit design techniques, provides a cost-effective alternative to bipolar and fast NMOS memories.

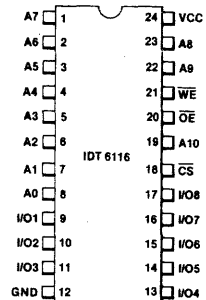
Access times as fast as 70ns are available with maximum power consumption of only 550mW. The circuit also offers a reduced power standby mode. When CS goes high, the circuit will automatically go to, and remain in, a standby power mode as long as CS remains high. In the standby mode, the low power device consumes less than 20μW typically. This capability provides significant system level power and cooling savings. Both versions also offer a battery backup data retention capability where the circuit typically consumes only 1 μW to 4 μW operating off of a 2V battery.

All inputs and outputs of the IDT6116 are TTL compatible and operation is from a single 5V supply, simplifying system designs. Fully static asynchronous circuitry is used, requiring no clocks or refreshing for operation, providing equal access and cycle times for ease of use.

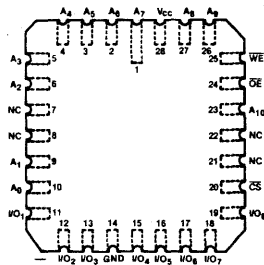
The IDT6116 is packaged in either a 24-pin, 600 mil or 300 mil DIP, 32-pin and 28-pin leadless chip carriers, or 24-pin flatpack, providing high board-level packing densities.

The IDT6116 Military RAM is 100% processed in compliance to the test methods of MIL-STD-883, Method 5004, making it ideally suited to military temperature applications demanding the highest level of performance and reliability.

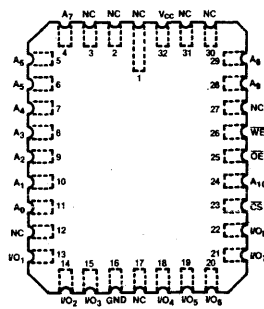
### PIN CONFIGURATIONS



24 PIN SIDEBRAZE TOP VIEW

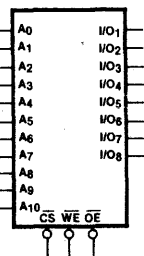


28 PIN LCC TOP VIEW



32 PIN LCC TOP VIEW

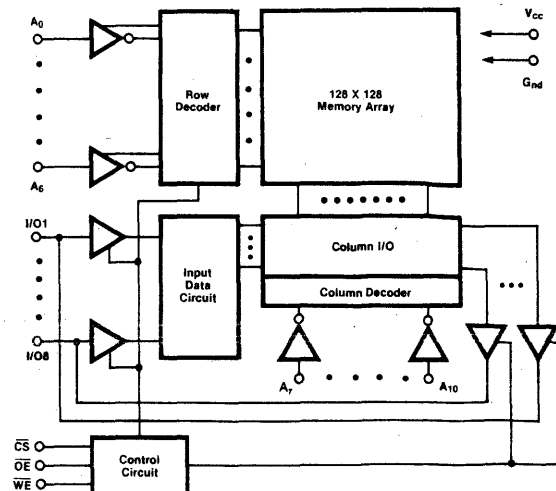
### LOGIC SYMBOL



### PIN NAMES

A <sub>0</sub> -A <sub>10</sub>	ADDRESS	WE	WRITE ENABLE
I/O <sub>1</sub> -I/O <sub>8</sub>	DATA INPUT/OUTPUT	OE	OUTPUT ENABLE
CS	CHIP SELECT	GND	GROUND
V <sub>CC</sub>	POWER		

### FUNCTIONAL BLOCK DIAGRAM



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MEMORY

**ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

TEMPERATURE RANGE		-55°C to +125°C	-40°C to +85°	0°C to +70°C	UNIT
SYMBOL	PARAMETER	RATING			
V <sub>TERM</sub>	Voltage on any Pin with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	-0.5 to +7.0	V
T <sub>A</sub>	Operating Temperature	-55 to +125	-40 to +85	0 to +70	°C
T <sub>BIAS</sub>	Temperature Under Bias	-65 to +135	-55 to +125	-10 to +85	°C
T <sub>STG</sub>	Storage Temperature	-65 to +150	-65 to +150	-55 to +125	°C
P <sub>T</sub>	Power Dissipation	1.0	1.0	1.0	W
I <sub>OUT</sub>	DC Output Current	50	50	50	mA

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum ratings conditions for extended periods may affect reliability.

**RECOMMENDED OPERATING CONDITIONS**

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
V <sub>CC</sub>	Supply Voltage	4.5	5.0	5.5	V
GND	Supply Voltage	0	0	0	V
V <sub>IH</sub>	Input High Voltage	2.2	3.5	6.0	V
V <sub>IL</sub>	Input Low Voltage	-1.0*	—	+0.8	V
C <sub>L</sub>	Output Load	—	—	100	pF
TTL	Output Load	—	—	1	—

MILITARY (T<sub>A</sub> = -55°C to +125°C)  
 INDUSTRIAL (T<sub>A</sub> = -40°C to +85°C)  
 COMMERCIAL (T<sub>A</sub> = 0°C to +70°C)

\* V<sub>IL</sub> min. = -3.5V for pulse widths less than 30 ns and duty cycles less than 50%.

**IDT6116S DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub>=5V ±10%, T<sub>A</sub>= -55°C to +125°C)**

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6116S90/120			IDT6116S150			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
I <sub>LI</sub>	Input Leakage Current	V <sub>CC</sub> = 5.5V, V <sub>IN</sub> = GND to V <sub>CC</sub>	—	—	10	—	—	10	μA
I <sub>LO</sub>	Output Leakage Current	$\overline{CS} = V_{IH}$ or $\overline{OE} = V_{IH}$ V <sub>I/O</sub> = GND to V <sub>CC</sub>	—	—	10	—	—	10	μA
I <sub>CC</sub>	Operating Power Supply Current	$\overline{CS} = V_{IL}$ , I <sub>I/O</sub> = 0mA	—	40	100	—	35	90	mA
I <sub>CC1</sub>		V <sub>IH</sub> = 3.5V, V <sub>IL</sub> = 0.6V I <sub>I/O</sub> = 0mA	—	35	—	—	30	—	mA
I <sub>CC2</sub>	Dynamic Operating Current	Min. Duty Cycle = 100%	—	40	100	—	35	90	mA
I <sub>SB</sub>	Standby Power Current	$\overline{CS} = V_{IH}$	—	5	25	—	5	25	mA
I <sub>SB1</sub>		$\overline{CS} \geq V_{CC} - 0.2V$ , V <sub>IN</sub> ≥ V <sub>CC</sub> - 0.2V or V <sub>IN</sub> ≤ 0.2V	—	0.02	10	—	0.02	10	mA
V <sub>OL</sub>	Output Low Voltage	I <sub>OL</sub> = 3.5mA <sup>(2)</sup>	—	—	0.4	—	—	0.4	V
V <sub>OH</sub>	Output High Voltage	I <sub>OH</sub> = -1.0mA	2.4	—	—	2.4	—	—	V

1. V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C  
 2. I<sub>OL</sub> = 2.1mA for IDT6116S150

**IDT6116L DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub>=5V ±10%, T<sub>A</sub>= -55°C to +125°C)**

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6116L90/120			IDT6116L150			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
I <sub>LI</sub>	Input Leakage Current	V <sub>CC</sub> = 5.5V, V <sub>IN</sub> = GND to V <sub>CC</sub>	—	—	5	—	—	5	μA
I <sub>LO</sub>	Output Leakage Current	$\overline{CS} = V_{IH}$ or $\overline{OE} = V_{IH}$ V <sub>I/O</sub> = GND to V <sub>CC</sub>	—	—	5	—	—	5	μA
I <sub>CC</sub>	Operating Power Supply Current	$\overline{CS} = V_{IL}$ , I <sub>I/O</sub> = 0mA	—	40/35	100/90	—	30	80	mA
I <sub>CC1</sub>		V <sub>IH</sub> = 3.5V, V <sub>IL</sub> = 0.6V I <sub>I/O</sub> = 0mA	—	35/30	—	—	30	—	mA
I <sub>CC2</sub>	Dynamic Operating Current	Min. Duty Cycle = 100%	—	40/35	100/90	—	30	80	mA
I <sub>SB</sub>	Standby Power Current	$\overline{CS} = V_{IH}$	—	5	20	—	5	15	mA
I <sub>SB1</sub>		$\overline{CS} \geq V_{CC} - 0.2V$ , V <sub>IN</sub> ≥ V <sub>CC</sub> - 0.2V or V <sub>IN</sub> ≤ 0.2V	—	4	900	—	4	900	μA
V <sub>OL</sub>	Output Low Voltage	I <sub>OL</sub> = 3.5mA <sup>(2)</sup>	—	—	0.4	—	—	0.4	V
V <sub>OH</sub>	Output High Voltage	I <sub>OH</sub> = -1.0mA	2.4	—	—	2.4	—	—	V

1. V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C  
 2. I<sub>OL</sub> = 2.1mA for IDT6116L150

**IDT6116S DC ELECTRICAL CHARACTERISTICS** ( $V_{CC}=5V \pm 10\%$ ,  $T_A = -40^\circ C$  to  $+85^\circ C$ )

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6116S90/120			IDT6116S150			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
$ I_{LI} $	Input Leakage Current	$V_{CC} = 5.5V, V_{IN} = GND \text{ to } V_{CC}$	—	—	10	—	—	10	$\mu A$
$ I_{LO} $	Output Leakage Current	$\overline{CS} = V_{IH} \text{ or } \overline{OE} = V_{IH}$ $V_{IO} = GND \text{ to } V_{CC}$	—	—	10	—	—	10	$\mu A$
$I_{CC}$	Operating Power Supply Current	$\overline{CS} = V_{IL}, I_{IO} = 0mA$	—	40	90	—	35	80	mA
$I_{CC1}$		$V_{IH} = 3.5V, V_{IL} = 0.6V$ $I_{IO} = 0mA$	—	35	—	—	30	—	mA
$I_{CC2}$	Dynamic Operating Current	Min. Duty Cycle = 100%	—	40	90	—	35	80	mA
$I_{SB}$	Standby Power Current	$\overline{CS} = V_{IH}$	—	5	20	—	5	20	mA
$I_{SB1}$		$\overline{CS} \geq V_{CC} - 0.2V,$ $V_{IN} \geq V_{CC} - 0.2V \text{ or } V_{IN} \leq 0.2V$	—	20	2000	—	20	2000	$\mu A$
$V_{OL}$	Output Low Voltage	$I_{OL} = 4mA^{(2)}$	—	—	0.4	—	—	0.4	V
$V_{OH}$	Output High Voltage	$I_{OH} = -1.0mA$	2.4	—	—	2.4	—	—	V

1.  $V_{CC} = 5V, T_A = 25^\circ C$   
2.  $I_{OL} = 2.1mA$  for IDT6116S150

**IDT6116L DC ELECTRICAL CHARACTERISTICS** ( $V_{CC}=5V \pm 10\%$ ,  $T_A = -40^\circ C$  to  $+85^\circ C$ )

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6116L90/120			IDT6116L150			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
$ I_{LI} $	Input Leakage Current	$V_{CC} = 5.5V, V_{IN} = GND \text{ to } V_{CC}$	—	—	2	—	—	2	$\mu A$
$ I_{LO} $	Output Leakage Current	$\overline{CS} = V_{IH} \text{ or } \overline{OE} = V_{IH}$ $V_{IO} = GND \text{ to } V_{CC}$	—	—	2	—	—	2	$\mu A$
$I_{CC}$	Operating Power Supply Current	$\overline{CS} = V_{IL}, I_{IO} = 0mA$	—	40/35	90/80	—	30	70	mA
$I_{CC1}$		$V_{IH} = 3.5V, V_{IL} = 0.6V$ $I_{IO} = 0mA$	—	35/30	—	—	30	—	mA
$I_{CC2}$	Dynamic Operating Current	Min. Duty Cycle = 100%	—	40/35	90/80	—	30	70	mA
$I_{SB}$	Standby Power Current	$\overline{CS} = V_{IH}$	—	5	20/15	—	4	12	mA
$I_{SB1}$		$\overline{CS} \geq V_{CC} - 0.2V,$ $V_{IN} \geq V_{CC} - 0.2V \text{ or } V_{IN} \leq 0.2V$	—	4	200	—	4	200	$\mu A$
$V_{OL}$	Output Low Voltage	$I_{OL} = 4mA^{(2)}$	—	—	0.4	—	—	0.4	V
$V_{OH}$	Output High Voltage	$I_{OH} = -1.0mA$	2.4	—	—	2.4	—	—	V

1.  $V_{CC} = 5V, T_A = 25^\circ C$   
2.  $I_{OL} = 2.1mA$  for IDT6116L150

**IDT6116S DC CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $Gnd = 0V, T_A = 0^\circ C$  to  $+70^\circ C$ )

SYMBOL	ITEM	TEST CONDITIONS	IDT6116S 70			IDT6116S 90/120			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP.*	MAX.	
$ I_{LI} $	Input Leakage Current	$V_{CC} = 5.5V, V_{IN} = GND \text{ to } V_{CC}$	—	—	10	—	—	10	$\mu A$
$ I_{LO} $	Output Leakage Current	$\overline{CS} = V_{IH} \text{ or } \overline{OE} = V_{IH}$ $V_{IO} = GND \text{ to } V_{CC}$	—	—	10	—	—	10	$\mu A$
$I_{CC}$	Operating Power Supply Current	$\overline{CS} = V_{IL}, I_{IO} = 0mA$	—	50	100	—	40	80	mA
$I_{CC1}$		$V_{IH} = 3.5V, V_{IL} = 0.6V$ $I_{IO} = 0mA$	—	40	—	—	35	—	mA
$I_{CC2}$	Average Operating Current	Min. Cycle Duty = 100%	—	50	100	—	40	80	mA
$I_{SB}$	Standby Power Current	$\overline{CS} = V_{IH}$	—	5	15	—	5	15	mA
$I_{SB1}$		$\overline{CS} \geq V_{CC} - 0.2V,$ $V_{IN} \geq V_{CC} - 0.2V \text{ or } V_{IN} \leq 0.2V$	—	20	2000	—	20	2000	$\mu A$
$V_{OL}$	Output Voltage	$I_{OL} = 4mA$	—	—	0.4	—	—	0.4	V
$V_{OH}$		$I_{OH} = -1.0mA$	2.4	—	—	2.4	—	—	V

1.  $V_{CC} = 5V, T_A = 25^\circ C$

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**IDT6116L DC CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ , Gnd = 0V,  $T_A = 0^\circ C$  to  $+70^\circ C$ )

SYMBOL	ITEM	TEST CONDITIONS	IDT6116L 90			IDT6116L 120/150			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
$ I_{L1} $	Input Leakage Current	$V_{CC} = 5.5V, V_{IN} = GND \text{ to } V_{CC}$	—	—	2	—	—	2	$\mu A$
$ I_{L0} $	Output Leakage Current	$\overline{CS} = V_{IH} \text{ or } \overline{OE} = V_{IH}$ $V_{I/O} = GND \text{ to } V_{CC}$	—	—	2	—	—	2	$\mu A$
$I_{CC}$	Operating Power Supply Current	$\overline{CS} = V_{IL}, I_{I/O} = 0mA$	—	40	80	—	35/30	70/60	mA
$I_{CC1}$		$V_{IH} = 3.5V, V_{IL} = 0.6V$ $I_{I/O} = 0mA$	—	35	—	—	30	—	mA
$I_{CC2}$	Average Operating Current	Min. Cycle Duty = 100%	—	40	80	—	35/30	70/60	mA
$I_{SB}$	Standby Power Current	$\overline{CS} = V_{IH}$	—	5	15	—	4	12	mA
$I_{SB1}$		$\overline{CS} \geq V_{CC} - 0.2V,$ $V_{IN} \geq V_{CC} - 0.2V \text{ or } V_{IN} \leq 0.2V$	—	4	100	—	4	100	$\mu A$
$V_{OL}$	Output Voltage	$I_{OL} = 4mA^{(2)}$	—	—	0.4	—	—	0.4	V
$V_{OH}$		$I_{OH} = -1.0mA$	2.4	—	—	2.4	—	—	V

1.  $V_{CC} = 5V, T_A = 25^\circ C$       2.  $I_{OL} = 2.1mA$  for IDT6116L 150

**AC ELECTRICAL CHARACTERISTICS**

( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C / -40^\circ C$  to  $+85^\circ C / 0^\circ C$  to  $+70^\circ C$ )<sup>(1)</sup>

SYMBOL	PARAMETER	<sup>(2)</sup> IDT616S70		IDT6116S90 IDT6116L90		IDT6116S120 IDT6116L120		IDT6116S150 IDS6116L150		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>										
$t_{RC}$	Read Access Time	70	—	90	—	120	—	150	—	ns
$t_{AA}$	Address Access Time	—	70	—	90	—	120	—	150	ns
$t_{ACS}$	Chip Select Access Time	—	70	—	90	—	120	—	150	ns
$t_{CLZ}$	Chip Selection to Output in Low Z	5	—	0/0/5	—	5/5/10	—	10/15 <sup>(3)</sup>	—	ns
$t_{OE}$	Output Enable to Output Valid	—	50	—	65	—	80	—	100	ns
$t_{OLZ}$	Output Enable to Output in Low Z	5	—	0/0/5	—	5/5/10	—	10/15 <sup>(3)</sup>	—	ns
$t_{CHZ}$	Chip Deselection to Output in High Z	0	35	0	40	0	40	0	50	ns
$t_{OHZ}$	Output Disable to Output in High Z	0	35	0	40	0	40	0	50	ns
$t_{OH}$	Output Hold from Address Change	5	—	0/0/5	—	5/5/10	—	10/15 <sup>(3)</sup>	—	na
<b>WRITE CYCLE</b>										
$t_{WC}$	Write Cycle Time	70	—	90	—	120	—	150	—	ns
$t_{CW}$	Chip Selection to End of Write	40	—	55	—	70	—	90	—	ns
$t_{AW}$	Address Valid to End of Write	65	—	80	—	105	—	120	—	ns
$t_{AS}$	Address Setup Time	15	—	15	—	20	—	20	—	ns
$t_{WP}$	Write Pulse Width	40	—	55	—	70	—	90	—	ns
$t_{WR}$	Write Recovery Time	5	—	10/5/5	—	10/5/5	—	10	—	ns
$t_{OHZ}$	Output Disable to Output in High Z	0	35	0	40	0	40	0	50	ns
$t_{WHZ}$	Write to Output in High Z	0	40	0	50	0	50	0	60	ns
$t_{DW}$	Data to Write Time Overlap	30	—	30	—	35	—	40	—	ns
$t_{DH}$	Data Hold from Write Time	5	—	10/10/5	—	10/10/5	—	10	—	ns
$t_{OW}$	Output Active from End of Write	0	—	0	—	0/5/5	—	5/10 <sup>(4)</sup>	—	ns

1. Parameters listing three limits apply in this Temp. Range order: Military/Industrial/Commercial. All other limits apply to all three Temp. Ranges.  
 2. IDT6116S70 available in Commercial  $0^\circ C$  to  $70^\circ C$  only.  
 3. 10ns applies over  $-55^\circ C$  to  $+125^\circ C$  and  $-40^\circ C$  to  $+85^\circ C$ ; 15ns applies over  $0^\circ C$  to  $+70^\circ C$ .  
 4.  $t_{OW} = 5ns$  over  $-55^\circ C$  to  $+125^\circ C$ ;  $t_{OW} = 10ns$  over  $-40^\circ C$  to  $+85^\circ C$  and  $0^\circ C$  to  $+70^\circ C$ .

**AC TEST CONDITIONS**

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	10ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	1 TTL Gate and $C_L = 100pF$ (including scope and jig)

**TRUTH TABLE**

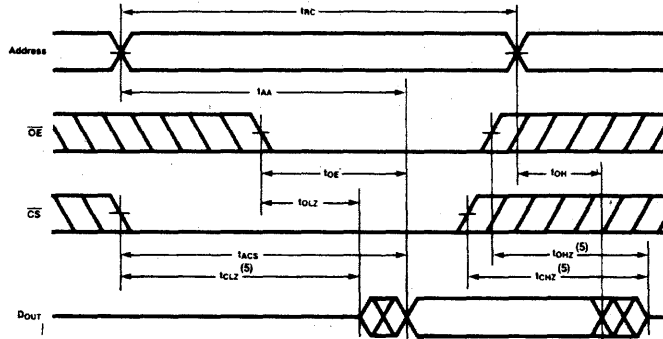
MODE	$\overline{CS}$	$\overline{OE}$	$\overline{WE}$	I/O OPERATION
Standby	H	X	X	High Z
Read	L	L	H	$D_{OUT}$
Read	L	H	H	High Z
Write	L	X	L	$D_{IN}$

**CAPACITANCE<sup>(1)</sup> ( $T_A = 25^\circ C, f = 1.0 MHz$ )**

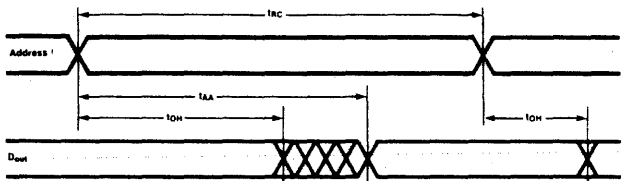
SYMBOL	PARAMETER	CONDITIONS	MAX.	UNIT
$C_{IN}$	Input Capacitance	$V_{IN} = 0V$	6	pF
$C_{I/O}$	Input/Output Capacitance	$V_{I/O} = 0V$	8	pF

1. This parameter is sampled and not 100% tested.

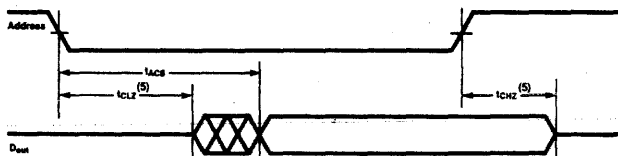
**TIMING WAVEFORMS OF READ CYCLE NO. 1<sup>(1)</sup>**



**READ CYCLE 2<sup>(1,2,4)</sup>**

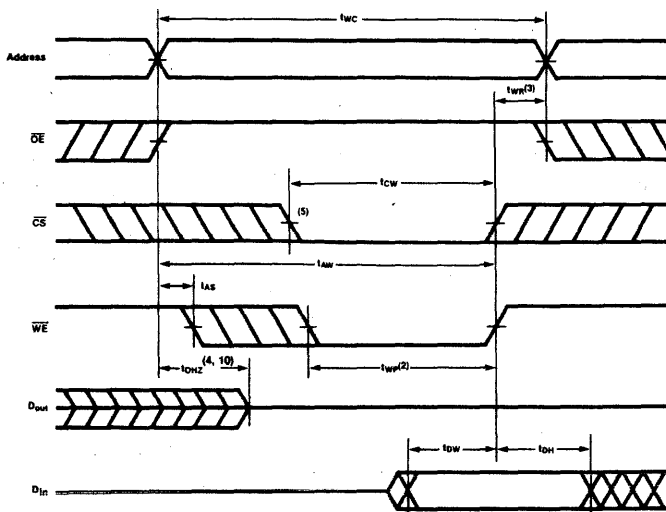


**READ CYCLE 3<sup>(1,3,4)</sup>**



- NOTES:
- $\overline{WE}$  is High for Read Cycle.
  - Device is continuously selected,  $\overline{CS} = V_{IL}$ .
  - Address valid prior to or coincident with  $\overline{CS}$  transition low.
  - $\overline{OE} = V_{IL}$ .
  - Transition is measured  $\pm 500mV$  from steady state. This parameter is sampled and not 100% tested.

**TIMING WAVEFORMS OF WRITE CYCLE 1<sup>(1)</sup>**

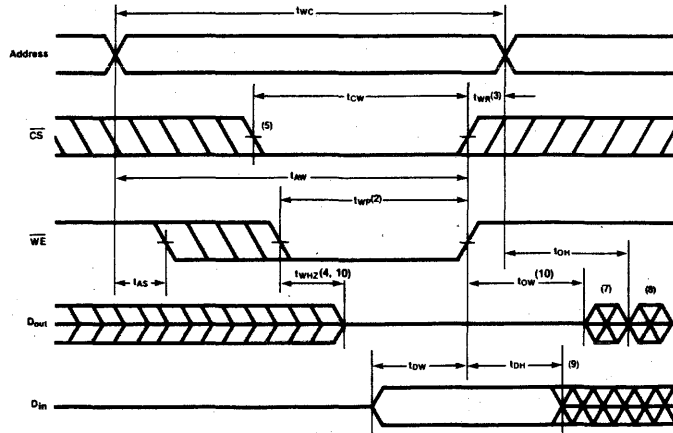


NOTE: See footnotes on next page.

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**TIMING WAVEFORM OF WRITE CYCLE 2<sup>(1,6)</sup>**



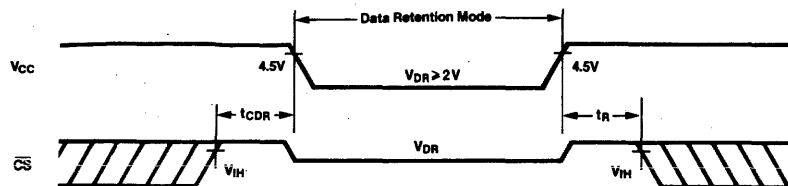
- NOTES: 1. WE must be high during all address transitions.  
 2. A write occurs during the overlap ( $t_{WP}$ ) of a low CS and a low WE.  
 3.  $t_{WR}$  is measured from the earlier of CS or WE going high to the end of write cycle.  
 4. During this period, I/O pins are in the output state so that the input signals of opposite phase to the outputs must not be applied.  
 5. If the CS low transition occurs simultaneously with the WE low transitions or after the WE transition, outputs remain in a high impedance state.  
 6. OE is continuously low ( $\overline{OE} = V_{IL}$ ).  
 7. D<sub>OUT</sub> is the same phase of write data of this write cycle.  
 8. D<sub>OUT</sub> is the read data of next address.  
 9. If CS is low during this period, I/O pins are in the output state. Then the data input signals of opposite phase to the outputs must not be applied to them.  
 10. Transition is measured  $\pm 500\text{mV}$  from steady state. This parameter is sampled and not 100% tested.

**DATA RETENTION CHARACTERISTICS**  
 ( $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  /  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$  /  $0^\circ\text{C}$  to  $70^\circ\text{C}$ )

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6116L			IDT6116S			UNIT	
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.		
V <sub>DR</sub>	V <sub>CC</sub> for Retention Data		2.0	—	—	2.0	—	—	V	
I <sub>CCDR</sub>	Data Retention Current	V <sub>CC</sub> = 2.0V, CS ≥ V <sub>CC</sub> - 0.2V V <sub>IN</sub> ≥ V <sub>CC</sub> - 0.2V or V <sub>IN</sub> ≤ 0.2V	COM'L	—	0.5	20	—	2	400	μA
			IND	—	0.5	40	—	2	600	μA
			MIL	—	0.5	300	—	2	1000	μA
t <sub>CDR</sub>	Chip Deselect to Data Retention Time		0	—	—	0	—	—	ns	
t <sub>r</sub>	Operation Recovery Time		t <sub>rc</sub> <sup>(2)</sup>	—	—	t <sub>rc</sub> <sup>(2)</sup>	—	—	ns	

1. V<sub>CC</sub> = 2V, T<sub>A</sub> = +25°C  
 2. t<sub>rc</sub> = Read Cycle Time

**LOW V<sub>CC</sub> DATA RETENTION WAVEFORM**



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Integrated Device Technology Inc.

# CMOS STATIC RAMS 16K (16K x 1 BIT)

# IDT6167S IDT6167L

MILITARY / INDUSTRIAL / COMMERCIAL TEMPERATURE RANGES

### FEATURES:

- High-speed (equal access and cycle time)
  - MILITARY/INDUSTRIAL
    - IDT6167S 55/70/85/100ns (max.)
    - IDT6167L 55/70/85/100ns (max.)
  - COMMERCIAL
    - IDT6167S 45/55/70/85ns (max.)
    - IDT6167L 45/55/70/85ns (max.)
- Low power consumption
 

IDT6167S	IDT6167L
Active: 150mW (typ.)	Active: 125mW (typ.)
Standby: 100µW (typ.)	Standby: 10µW (typ.)
- Battery backup operation — 2V data retention voltage (IDT6167L only)
- High-density 20-pin dual-in-line package and 20-pin leadless chip carriers
- Produced with advanced CEMOST<sup>TM</sup> I high-performance technology
- CEMOST<sup>TM</sup> I process virtually eliminates alpha particle soft-error rates (with no organic die coatings)
- Separate data input and output
- Single 5V (±10%) power supply
- Input and output directly TTL-compatible
- Three-state output
- Static operation: no clocks or refresh required
- Pin-compatible with standard 16K x 1 static RAMs
- Standard product 100% screened to MIL-STD-883 Class C
- Military product available 100% screened to Class B

### DESCRIPTION:

The IDT6167 is a 16,384-bit high-speed static RAM organized as 16K x 1. It is fabricated using IDT's high-performance, high-reliability technology — CEMOST<sup>TM</sup> I. This state-of-the-art technology, combined with innovative circuit design techniques, provides a cost effective alternative to bipolar and fast NMOS memories.

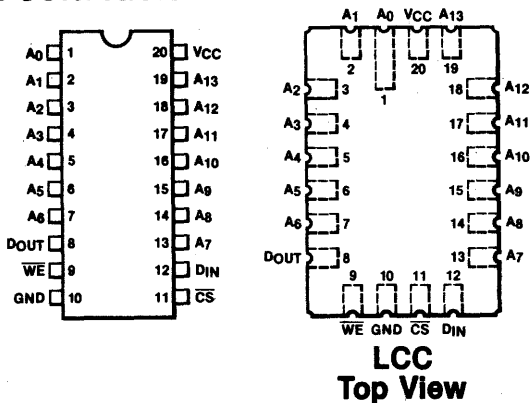
Access times as fast as 45ns are available with maximum power consumption of only 330mW. The circuit also offers a reduced power standby mode. When CS goes high, the circuit will automatically go to, and remain in, a standby mode as long as CS remains high. In the standby mode, the device consumes less than 100µW, typically. This capability provides significant system-level power and cooling savings. The low power, (L), version also offers a battery backup data retention capability where the circuit typically consumes only 1µW operating off of a 2V battery.

All inputs and the output of the IDT6167 are TTL-compatible and operate from a single 5V supply, thus simplifying system designs. Fully static asynchronous circuitry is used, which requires no clocks or refreshing for operation, and provides equal access and cycle times for ease of use.

The IDT6167 is packaged in either a space-saving 20-pin, 300 mil DIP, a 20-pin leadless chip carrier, or 20-pin flatpack providing high board-level packing densities.

The IDT6167 Military RAM is 100% processed in compliance to the test methods of MIL-STD-883, Method 5004, making it ideally suited to military temperature applications demanding the highest level of performance and reliability.

### PIN CONFIGURATIONS

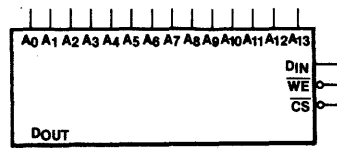


### PIN NAMES

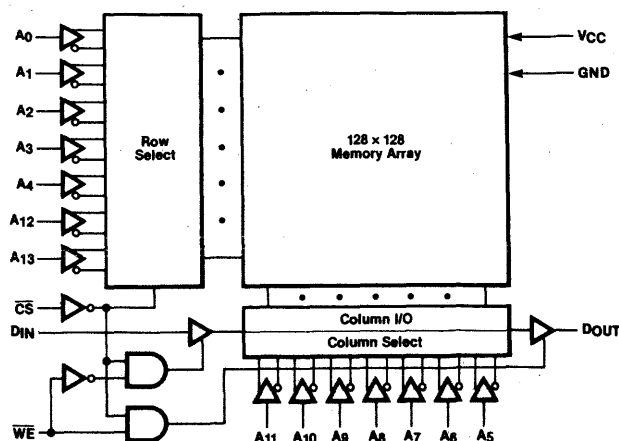
A <sub>0</sub> -A <sub>13</sub>	ADDRESS INPUTS	D <sub>IN</sub>	DATA IN
CS	CHIP SELECT	D <sub>OUT</sub>	DATA OUT
WE	WRITE ENABLE	GND	GROUND
V <sub>CC</sub>	POWER		

CEMOS is a trademark of Integrated Device Technology, Inc.

### LOGIC SYMBOL



### FUNCTIONAL BLOCK DIAGRAM





**ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

TEMPERATURE RANGE		-55°C to +125°	-40°C to +85°C	0°C to +70°C	UNIT
SYMBOL	RATING	VALUE			
V <sub>TERM</sub>	Terminal Voltage with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	-0.5 to +7.0	V
T <sub>A</sub>	Operating Temperature	-55 to +125	-40 to +85	0 to +70	°C
T <sub>BIAS</sub>	Temperature Under Bias	-65 to +135	-55 to +125	-10 to +85	°C
T <sub>STG</sub>	Storage Temperature	-65 to +150	-65 to +150	-55 to +125	°C
P <sub>T</sub>	Power Dissipation	1.0	1.0	1.0	W
I <sub>OUT</sub>	DC Output Current	20	20	20	mA

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

**RECOMMENDED DC OPERATING CONDITIONS**

(T<sub>A</sub> = -55°C to +125°C / -40°C to +85°C / 0°C to +70°C)

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
V <sub>CC</sub>	Supply Voltage	4.5	5.0	5.5	V
GND	Supply Voltage	0	0	0	V
V <sub>IH</sub>	Input High Voltage	2.2	—	6.0	V
V <sub>IL</sub>	Input Low Voltage	-0.5*	—	0.8	V

MILITARY (T<sub>A</sub> = -55°C to +125°C)

INDUSTRIAL (T<sub>A</sub> = -40°C to +85°C)

COMMERCIAL (T<sub>A</sub> = 0°C to +70°C)

\* V<sub>IL</sub> min. = -3.5V for pulse widths less than 30ns and duty cycles less than 50%.

**DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub> = 5V ± 10%, T<sub>A</sub> = -55°C to +125°C / -40°C to +85°C)**

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6167S			IDT6167L			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
I <sub>LI</sub>	Input Leakage Current	V <sub>CC</sub> = 5.5V, V <sub>IN</sub> = 0V to V <sub>CC</sub>	—	—	10	—	—	5/2	μA
I <sub>LO</sub>	Output Leakage Current	CS = V <sub>IH</sub> , V <sub>OUT</sub> = 0V to V <sub>CC</sub>	—	—	10	—	—	5/2	μA
I <sub>CC1</sub>	Operating Power Supply Current	CS = V <sub>IL</sub> , Output Open	—	30	60	—	25	50	mA
I <sub>CC2</sub>	Dynamic Operating Current	Min. Duty Cycle = 100%	—	30	60	—	25	50	mA
I <sub>SB</sub>	Standby Power Supply Current	CS ≥ V <sub>IH</sub>	—	5	20	—	5	20	mA
I <sub>SB1</sub>	Full Standby Power Supply Current	CS ≥ V <sub>CC</sub> - 0.2V V <sub>IN</sub> ≥ V <sub>CC</sub> - 0.2V or ≤ 0.2V	—	20	10,000/ 2000	—	2	900/ 150	μA
V <sub>OL</sub>	Output Low Voltage	I <sub>OL</sub> = 8mA	—	—	0.4	—	—	0.4	V
V <sub>OH</sub>	Output High Voltage	I <sub>OH</sub> = -4mA	2.4	—	—	2.4	—	—	V

1. V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C

**DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub> = 5V ± 10%, T<sub>A</sub> = 0 to +70°C)**

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6167S			IDT6167L			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
I <sub>LI</sub>	Input Leakage Current	V <sub>CC</sub> = 5.5V, V <sub>IN</sub> = 0V to V <sub>CC</sub>	—	—	2	—	—	2	μA
I <sub>LO</sub>	Output Leakage Current	CS = V <sub>IH</sub> , V <sub>OUT</sub> = 0V to V <sub>CC</sub>	—	—	2	—	—	2	μA
I <sub>CC1</sub>	Operating Power Supply Current	CS = V <sub>IL</sub> , Output Open	—	30	60	—	25	50	mA
I <sub>CC2</sub>	Dynamic Operating Current	Min. Duty Cycle = 100%	—	30	60	—	25	50	mA
I <sub>SB</sub>	Standby Power Supply Current	CS ≥ V <sub>IH</sub>	—	5	20	—	5	20	mA
I <sub>SB1</sub>	Full Standby Power Supply Current	CS ≥ V <sub>CC</sub> - 0.2V V <sub>IN</sub> ≥ V <sub>CC</sub> - 0.2V or ≤ 0.2V	—	20	2000	—	2	50	μA
V <sub>OL</sub>	Output Low Voltage	I <sub>OL</sub> = 8mA	—	—	0.4	—	—	0.4	V
V <sub>OH</sub>	Output High Voltage	I <sub>OH</sub> = -4mA	2.4	—	—	2.4	—	—	V

1. V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C

**TRUTH TABLE**

MODE	CS	WE	OUTPUT	POWER
Standby	H	X	High Z	Standby
Read	L	H	D Out	Active
Write	L	L	High Z	Active

**CAPACITANCE (T<sub>A</sub> = 25°C, f = 1.0 MHz)**

SYMBOL	ITEM	CONDITIONS	MAX.	UNIT
C <sub>IN</sub>	Input Capacitance	V <sub>IN</sub> = 0V	5	pF
C <sub>OUT</sub>	Output Capacitance	V <sub>OUT</sub> = 0V	6	pF

NOTE: This parameter is sampled and not 100% tested.

**AC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C / -40^\circ C$  to  $+85^\circ C$  unless otherwise noted.)

SYMBOL	PARAMETER	IDT6167S55 IDT6167L55		IDT6167S70 IDT6167L70		IDT6167S85 IDT6167L85		IDT6167S100 IDT6167L100		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>										
$t_{RC}$ (TAVAV)	Read Cycle Time	55	—	70	—	85	—	100	—	ns
$t_{AA}$ (TAVQV)	Address Access Time	—	55	—	70	—	85	—	100	ns
$t_{ACS}$ (TELQV)	Chip Select Access Time	—	55	—	70	—	85	—	100	ns
$t_{OH}$ (TAXQX)	Output Hold from Address Change	5	—	5	—	5	—	5	—	ns
$t_{LZ}$ (TELQX)	Chip Selection to Output in Low Z	5	—	5	—	5	—	5	—	ns
$t_{HZ}$ (TEHQZ)	Chip Deselection to Output in High Z	0	40	0	40	0	50	0	50	ns
$t_{PU}$ (TELICCH)	Chip Selection to Power Up Time	0	—	0	—	0	—	0	—	ns
$t_{PD}$ (TEHICCL)	Chip Deselection to Power Down Time	—	55	—	70	—	85	—	100	ns
<b>WRITE CYCLE</b>										
$t_{WC}$ (TAVAV)	Write Cycle Time	55	—	70	—	85	—	100	—	ns
$t_{CW}$ (TELWH)	Chip Selection to End of Write	45	—	55	—	65	—	80	—	ns
$t_{AW}$ (TAVWH)	Address Valid to End of Write	45	—	55	—	65	—	80	—	ns
$t_{AS}$ (TAVWL)	Address Setup Time	0	—	0	—	0	—	0	—	ns
$t_{WP}$ (TWLWH)	Write Pulse Width	35	—	40	—	45	—	55	—	ns
$t_{WR}$ (TWHAX)	Write Recovery Time	0	—	0	—	0	—	0	—	ns
$t_{DW}$ (TDVWH)	Data Valid to End of Write	25	—	30	—	35	—	40	—	ns
$t_{DH}$ (TWHDX)	Data Hold Time	0	—	0	—	0	—	0	—	ns
$t_{WZ}$ (TWLQZ)	Write Enable to Output in High Z	0	40	0	40	0	50	0	50	ns
$t_{OW}$ (TWHQX)	Output Active from End of Write	0	—	0	—	0	—	0	—	ns

**AC CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = 0$  to  $+70^\circ C$ ) unless otherwise noted.)

SYMBOL	PARAMETER	IDT6167S45 IDT6167L45		IDT6167S55 IDT6167L55		IDT6167S70 IDT6167L70		IDT6167S85 IDT6167L85		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>										
$t_{RC}$ (TAVAV)	Read Cycle Time	45	—	55	—	70	—	85	—	ns
$t_{AA}$ (TAVQV)	Address Access Time	—	45	—	55	—	70	—	85	ns
$t_{ACS}$ (TELQV)	Chip Select Access Time	—	45	—	55	—	70	—	85	ns
$t_{OH}$ (TAXQX)	Output Hold from Address Change	5	—	5	—	5	—	5	—	ns
$t_{LZ}$ (TELQX)	Chip Selection to Output in Low Z	5	—	5	—	5	—	5	—	ns
$t_{HZ}$ (TEHQZ)	Chip Deselection to Output in High Z	0	30	0	30	0	30	0	40	ns
$t_{PU}$ (TELICCH)	Chip Selection to Power Up Time	0	—	0	—	0	—	0	—	ns
$t_{PD}$ (TEHICCL)	Chip Deselection to Power Down Time	—	35	—	35	—	35	—	40	ns
<b>WRITE CYCLE</b>										
$t_{WC}$ (TAVAV)	Write Cycle Time	45	—	55	—	70	—	85	—	ns
$t_{CW}$ (TELWH)	Chip Selection to End of Write	40	—	45	—	55	—	65	—	ns
$t_{AW}$ (TAVWH)	Address Valid to End of Write	40	—	45	—	55	—	65	—	ns
$t_{AS}$ (TAVWL)	Address Setup Time	0	—	0	—	0	—	0	—	ns
$t_{WP}$ (TWLWH)	Write Pulse Width	30	—	35	—	40	—	45	—	ns
$t_{WR}$ (TWHAX)	Write Recovery Time	0	—	0	—	0	—	0	—	ns
$t_{DW}$ (TDVWH)	Data Valid to End of Write	25	—	25	—	30	—	35	—	ns
$t_{DH}$ (TWHDX)	Data Hold Time	0	—	0	—	0	—	0	—	ns
$t_{WZ}$ (TWLQZ)	Write Enable to Output in High Z	0	30	0	30	0	30	0	40	ns
$t_{OW}$ (TWHQX)	Output Active from End of Write	0	—	0	—	0	—	0	—	ns

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MEMORY

**AC TEST CONDITIONS**

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	5ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	See Figures 1 and 2

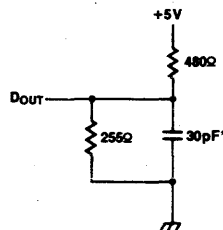


Figure 1. Output Load

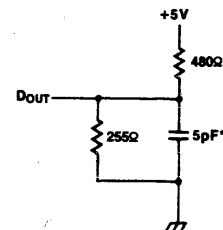
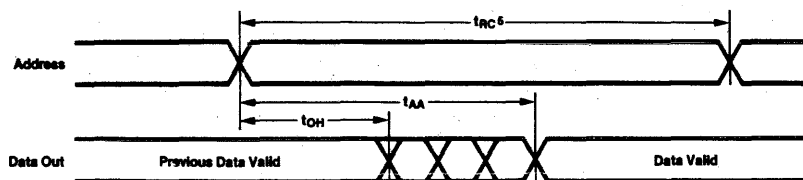


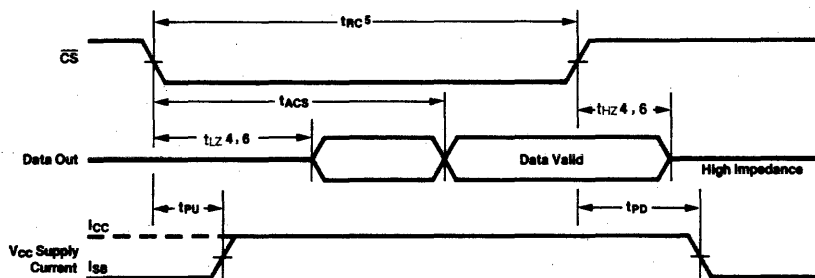
Figure 2. Output Load (for  $t_{HZ}$ ,  $t_{LZ}$ ,  $t_{WZ}$ , and  $t_{OZ}$ )

\*Including scope and jig.

**TIMING WAVEFORM OF READ CYCLE NO. 1<sup>(1,2)</sup>**

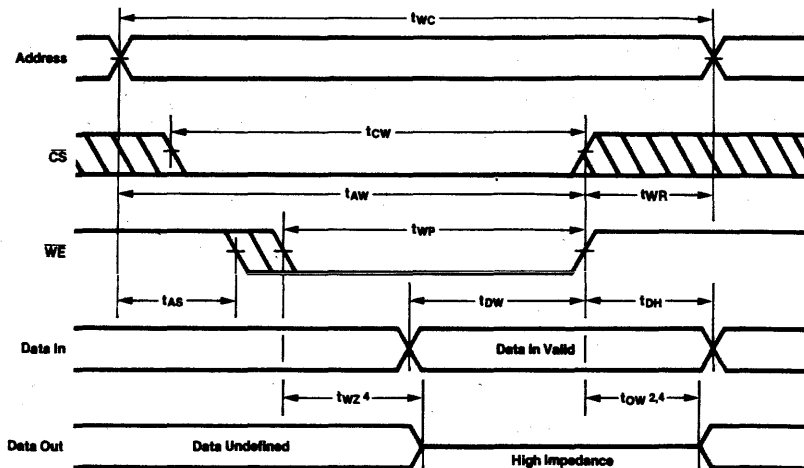


**TIMING WAVEFORM OF READ CYCLE NO. 2<sup>(1,3)</sup>**



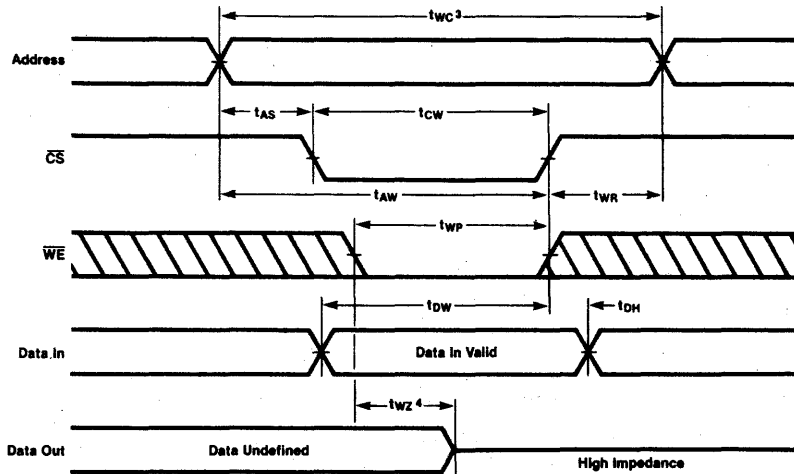
- NOTES:
1.  $\overline{WE}$  is high for READ cycle.
  2.  $\overline{CS}$  is low for READ cycle.
  3. Address valid prior to or coincident with  $\overline{CS}$  transition low.
  4. Transition is measured  $\pm 500mV$  from steady state voltage with specified loading in Figure 2. This parameter is sampled, not 100% tested.
  5. All READ cycle timings are referenced from the last valid address to the first transitioning address.
  6. For any given speed grade, operating voltage and temperature,  $t_{HZ}$  will be less than or equal to  $t_{LZ}$ .

**TIMING WAVEFORM OF WRITE CYCLE NO. 1 ( $\overline{WE}$  CONTROLLED)<sup>(1)</sup>**



- NOTES:
1.  $\overline{CS}$  or  $\overline{WE}$  must be high during address transitions.
  2. If  $\overline{CS}$  goes high simultaneously with  $\overline{WE}$  high, the output remains in a high impedance state.
  3. All write cycle timings are referenced from the last valid address to the first transitioning address.
  4. Transition is measured  $\pm 500mV$  from steady state voltage with specified loading in Figure 2. This parameter is sampled and not 100% tested.

**TIMING WAVEFORM OF WRITE CYCLE NO. 2 (CS CONTROLLED)<sup>(1)</sup>**



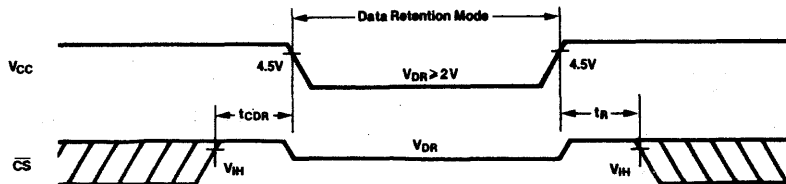
- NOTES: 1.  $\overline{CS}$  or  $\overline{WE}$  must be high during address transitions.  
 2. If  $\overline{CS}$  goes high simultaneously with  $\overline{WE}$  high, the output remains in a high impedance state.  
 3. All write cycle timings are referenced from the last valid address to the first transitioning address.  
 4. Transition is measured  $\pm 500\text{mV}$  from steady state voltage with specified loading in Figure 2. This parameter is sampled and not 100% tested.

**LOW  $V_{CC}$  DATA RETENTION CHARACTERISTICS FOR L VERSION ONLY ( $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  /  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$  /  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ )**

SYMBOL	PARAMETER	TEST CONDITION	MIN.	TYP. <sup>1</sup>	MAX.	UNIT	
$V_{DR}$	$V_{CC}$ for Data Retention		2.0	—	—	V	
$I_{CCDR}$	Data Retention Current	$\overline{CS} \geq V_{CC} - 0.2\text{V}$ $V_{IN} \geq V_{CC} - 0.2\text{V}$ or $\leq 0.2\text{V}$	COM'L	—	0.5 <sup>(2)</sup> 1.0 <sup>(3)</sup>	20 <sup>(2)</sup> 30 <sup>(2)</sup>	$\mu\text{A}$
			IND	—	0.5 <sup>(2)</sup> 1.0 <sup>(3)</sup>	40 <sup>(2)</sup> 60 <sup>(3)</sup>	
			MIL	—	0.5 <sup>(2)</sup> 1.0 <sup>(3)</sup>	300 <sup>(2)</sup> 450 <sup>(3)</sup>	
$t_{CDR}$	Chip Deselect to Data Retention Time		0	—	—	ns	
$t_R$	Operation Recovery Time		$t_{RC}$ <sup>(4)</sup>	—	—	ns	

- NOTES: 1.  $T_A = 25^\circ\text{C}$       3. at  $V_{CC} = 3\text{V}$   
 2. at  $V_{CC} = 2\text{V}$       4.  $t_{RC}$  = Read Cycle Time

**LOW  $V_{CC}$  DATA RETENTION WAVEFORM**



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MEMORY

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Integrated Device Technology, Inc.

# CMOS STATIC RAMS 16K (4K x 4 BIT)

# IDT6168S/L IDT71681S/L IDT71682S/L

## MILITARY / INDUSTRIAL / COMMERCIAL TEMPERATURE RANGES

### FEATURES:

- High-speed (equal access and cycle time)
  - IDT6168S/IDT6168L 45 (com'l)/55/70/85/100 ns (max.)
  - IDT71681S/IDT71681L 45 (com'l)/55/70/85/100 ns (max.)
  - IDT71682S/IDT71682L 45 (com'l)/55/70/85/100 ns (max.)
- Low power consumption
  - IDT6168S/IDT71681S/IDT71682S
    - Active: 225 mW (typ.)
    - Standby: 100  $\mu$ W (typ.)
  - IDT6168L/IDT71681L/IDT71682L
    - Active: 225 mW (typ.)
    - Standby: 10  $\mu$ W (typ.)
- Bidirectional data input and output (IDT6168 only)
- Separate data inputs and outputs (IDT71681 and IDT71682 only)
- Outputs in high impedance in WRITE mode (IDT71682 only)
- Battery backup operation — 2V data retention voltage (L versions only)
- Produced with advanced CEMOS™ I high-performance technology
- CEMOS I process virtually eliminates alpha particle soft-error rates (with no organic die coatings)
- Single 5V ( $\pm 10\%$ ) power supply
- Input and output directly TTL-compatible
- Three-state output
- Static operation: no clocks or refresh required
- Standard product 100% screened to MIL-STD-883 Class C
- Military product available 100% screened to Class B

### DESCRIPTION:

The IDT6168, IDT71681 and IDT71682 are 16,384-bit high-speed static RAMs organized as 4K x 4. They are

fabricated using IDT's high-performance, high-reliability technology — CEMOS™ I. This state-of-the-art technology, combined with innovative circuit design techniques, provides a cost effective alternative to bipolar and fast NMOS memories.

The IDT71681 and IDT71682 are manufactured with separate data inputs and outputs. This feature allows for improved system architectures. The external latches of a bidirectional bus system are not needed, improving both board density and system power levels. The absence of bus contention allows a streamlined approach, enhancing system speeds. Also, they fit readily into a pipeline structure.

Access times as fast as 45 ns are available with maximum power consumption of only 495 mW. The circuit also offers a reduced power standby mode. When CS goes high, the circuit will automatically go to, and remain in, a standby mode as long as CS remains high. In the standby mode, the device consumes less than 100  $\mu$ W, typically. This capability provides significant system-level power and cooling savings. The low power (L) version also offers a battery backup data retention capability where the circuit typically consumes only 1  $\mu$ W operating off a 2V battery.

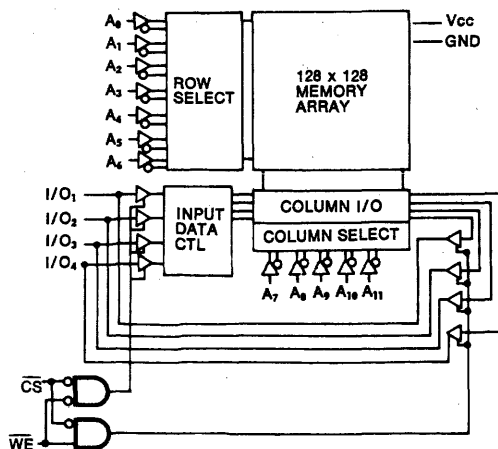
All inputs and outputs of the IDT6168, IDT71681 and IDT71682 are TTL-compatible and operate from a single 5V supply, thus simplifying system designs. Fully static asynchronous circuitry is used, which requires no clocks or refreshing for operation, and provides equal access and cycle times for ease of use.

The IDT6168 is packaged in either a space-saving 20-pin, 300 mil DIP, a 20-pin leadless chip carrier or a 20-pin flatpack and the IDT71681 and IDT71682 in a 24-pin, 300 mil DIP or a 28-pin leadless chip carrier, providing high board-level packing densities.

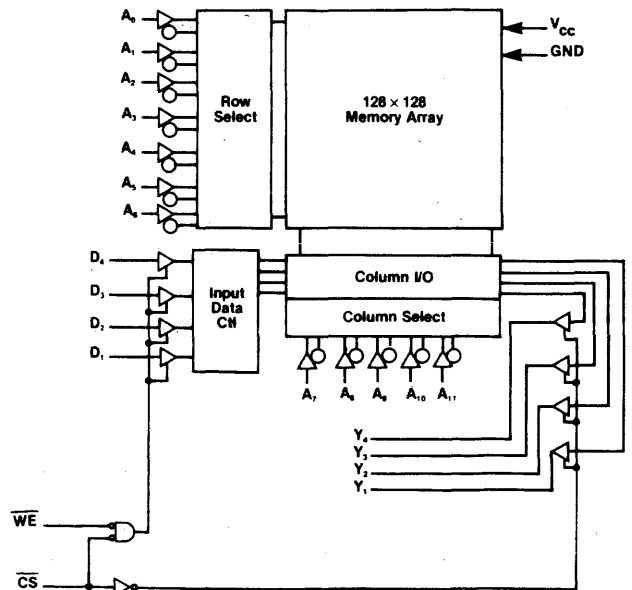
The Military RAMs are 100% processed in compliance to the test methods of MIL-STD-883, Method 5004, making them ideally suited to military temperature applications demanding the highest level of performance and reliability.

### FUNCTIONAL BLOCK DIAGRAMS

#### IDT6168



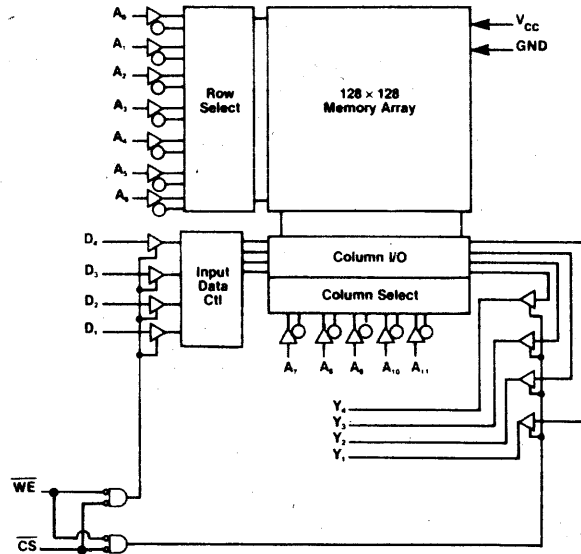
#### IDT71681



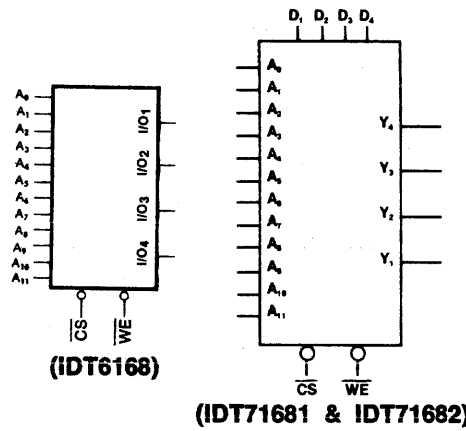
CEMOS is a trademark of Integrated Device Technology, Inc.

**FUNCTIONAL BLOCK DIAGRAMS**

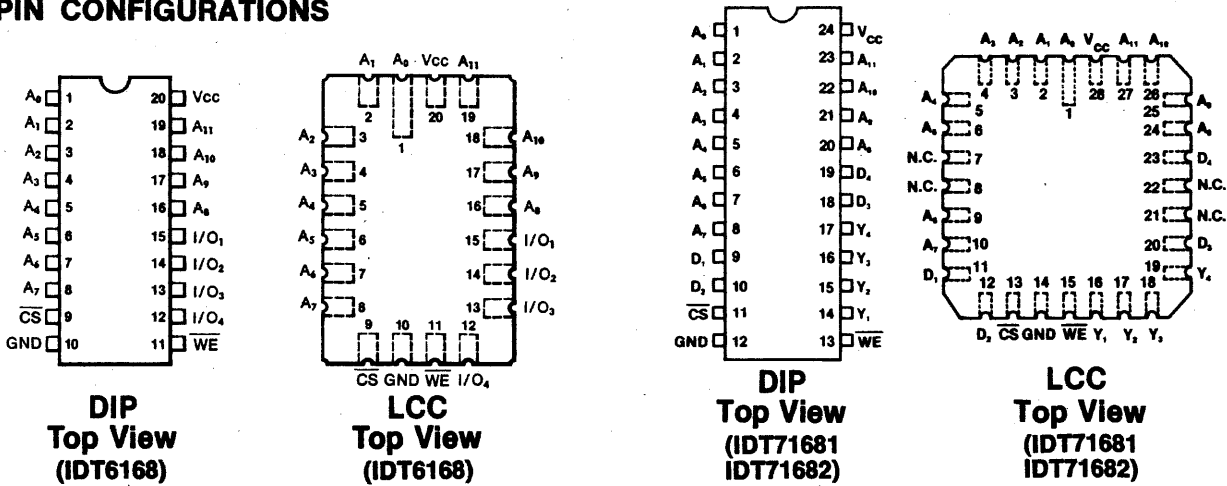
**IDT71682**



**LOGIC SYMBOLS**



**PIN CONFIGURATIONS**



**PIN NAMES (IDT6168)**

A <sub>0</sub> -A <sub>11</sub>	ADDRESS INPUTS	I/O <sub>1</sub> -I/O <sub>4</sub>	DATA INPUT/OUTPUT
CS	CHIP SELECT	V <sub>CC</sub>	POWER
WE	WRITE ENABLE	GND	GROUND

**(IDT71681 & IDT71682)**

A <sub>0</sub> -A <sub>11</sub>	ADDRESS INPUTS	D <sub>1</sub> -D <sub>4</sub>	DATA IN
CS	CHIP SELECT	Y <sub>1</sub> -Y <sub>4</sub>	DATA OUT
WE	WRITE ENABLE	GND	GROUND
V <sub>CC</sub>	POWER		

### ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

TEMPERATURE RANGE		-55°C to +125°C	-40°C to +85°	0°C to +70°C	UNIT
SYMBOL	PARAMETER	RATING			
V <sub>TERM</sub>	Voltage on any Pin with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	-0.5 to +7.0	V
T <sub>A</sub>	Operating Temperature	-55 to +125	-40 to +85	0 to +70	°C
T <sub>BIAS</sub>	Temperature Under Bias	-65 to +135	-55 to +125	-10 to +85	°C
T <sub>STG</sub>	Storage Temperature	-65 to +150	-65 to +150	-55 to +125	°C
P <sub>T</sub>	Power Dissipation	1.0	1.0	1.0	W
I <sub>OUT</sub>	DC Output Current	50	50	50	mA

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum ratings conditions for extended periods may affect reliability.

### RECOMMENDED DC OPERATING CONDITIONS

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
V <sub>CC</sub>	Supply Voltage	4.5	5.0	5.5	V
GND	Supply Voltage	0	0	0	V
V <sub>IH</sub>	Input High Voltage	2.2	3.5	6.0	V
V <sub>IL</sub>	Input Low Voltage	-1.0*	-	+0.8	V

MILITARY (T<sub>A</sub> = -55°C to +125°C)

INDUSTRIAL (T<sub>A</sub> = -40°C to +85°C)

COMMERCIAL (T<sub>A</sub> = 0°C to +70°C)

\* V<sub>IL</sub> min. = -3.5V for pulse widths less than 30 ns and duty cycles less than 50%.

### DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub> = 5V ± 10%, T<sub>A</sub> = -55°C to +125°C / -40°C to +85°C / 0°C to +70°C)

SYMBOL	PARAMETER	TEST CONDITIONS	IDT6168S/71681S			IDT6168L/71681L			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	MIN.	TYP. <sup>(1)</sup>	MAX.	
I <sub>LI</sub>	Input Leakage Current	V <sub>CC</sub> = 5.5V, V <sub>IN</sub> = 0V to V <sub>CC</sub>	-	-	10/10/2	-	-	5/2/2	μA
I <sub>LO</sub>	Output Leakage Current	$\overline{CS} = V_{IH}$ , V <sub>OUT</sub> = 0V to V <sub>CC</sub>	-	-	10/10/2	-	-	5/2/2	μA
I <sub>CC1</sub>	Operating Power Supply Current	$\overline{CS} = V_{IL}$ , Output Open	-	45	90	-	45	90	mA
I <sub>CC2</sub>	Dynamic Operating Current	Min. Duty Cycle = 100%	-	45	90	-	45	90	mA
I <sub>SB</sub>	Standby Power Supply Current	$\overline{CS} \geq V_{IH}$	-	5	20	-	5	20	mA
I <sub>SB1</sub>	Full Standby Power Supply Current	$\overline{CS} \geq V_{CC} - 0.2V$ V <sub>IN</sub> ≥ V <sub>CC</sub> - 0.2V or ≤ 0.2V	-	20	10,000/ 2000/2000	-	2	900/ 150/50	μA
V <sub>OL</sub>	Output Low Voltage	I <sub>OL</sub> = 8mA	-	-	0.4	-	-	0.4	V
V <sub>OH</sub>	Output High Voltage	I <sub>OH</sub> = -4mA	2.4	-	-	2.4	-	-	V

1. V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C

### TRUTH TABLE

MODE	$\overline{CS}$	$\overline{WE}$	OUTPUT	POWER
Standby	H	X	High Z	Standby
Read	L	H	D Out	Active
Write	L	L	High Z	Active

### CAPACITANCE (T<sub>A</sub> = 25°C, f = 1.0MHz)

SYMBOL	ITEM	CONDITIONS	MAX.	UNIT
C <sub>IN</sub>	Input Capacitance	V <sub>IN</sub> = 0V	5	pF
C <sub>OUT</sub>	Output Capacitance	V <sub>OUT</sub> = 0V	7	pF

NOTE: This parameter is sampled and not 100% tested.

### AC TEST CONDITIONS

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	5ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	See Figures 1 and 2

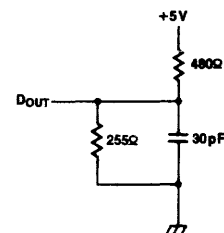


Figure 1. Output Load

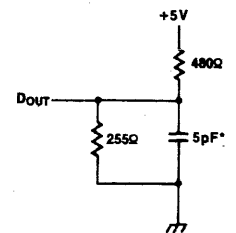


Figure 2. Output Load (for t<sub>HZ</sub>, t<sub>LZ</sub>, t<sub>wZ</sub>, and t<sub>OW</sub>)

\*Including scope and jig.

**AC CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$  /  $-40^\circ C$  to  $+85^\circ C$  /  $0^\circ C$  to  $+70^\circ C$  unless otherwise noted.)<sup>(1)</sup>

SYMBOL	PARAMETER	IDT6168S45 <sup>(2)</sup> IDT6168L45		IDT6168S55 IDT6168L55		IDT6168S70 IDT6168L70		IDT6168S85 IDT6168L85		IDT6168S100 <sup>(3)</sup> IDT6168L100		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>												
$t_{RC}$ (TAVAV)	Read Cycle Time	45	—	55	—	70	—	85	—	100	—	ns
$t_{AA}$ (TAVQV)	Address Access Time	—	45	—	55	—	70	—	85	—	100	ns
$t_{ACS}$ (TELOV)	Chip Select Access Time	—	45	—	55	—	70	—	85	—	100	ns
$t_{OH}$ (TAXQX)	Output Hold from Address Change	5	—	5	—	5	—	5	—	5	—	ns
$t_{LZ}$ (TELQX)	Chip Selection to Output in Low Z	20	—	20	—	20	—	20	—	20	—	ns
$t_{HZ}$ (TEHQZ)	Chip Deselection to Output in High Z	—	20	—	25	—	30/30/25	—	40/40/25	—	50	ns
$t_{PU}$ (TELICCH)	Chip Selection to Power Up Time	0	—	0	—	0	—	0	—	0	—	ns
$t_{PD}$ (TEHICCL)	Chip Deselection to Power Down Time	—	40	—	55/55/50	—	70/70/60	—	85/85/70	—	100	ns
$t_{RCS}$	Read Command Set-Up Time	-5	—	-5	—	-5	—	-5	—	-5	—	ns
$t_{RCH}$	Read Command Hold Time	-5	—	-5	—	-5	—	-5	—	-5	—	ns
<b>WRITE CYCLE</b>												
$t_{WC}$ (TAVAV)	Write Cycle Time	40	—	50	—	60	—	75	—	90	—	ns
$t_{CW}$ (TELWH)	Chip Selection to End of Write	35	—	50/50/45	—	60/60/55	—	75/75/65	—	90	—	ns
$t_{AW}$ (TAVWH)	Address Valid to End of Write	35	—	50/50/45	—	60/60/55	—	75/75/65	—	90	—	ns
$t_{AS}$ (TAVWL)	Address Setup Time	0	—	0	—	0	—	0	—	0	—	ns
$t_{WP}$ (TWLWH)	Write Pulse Width	35	—	50/50/45	—	60/60/55	—	75/75/70	—	90	—	ns
$t_{WR}$ (TWHAX)	Write Recovery Time	0	—	0	—	0	—	0	—	0	—	ns
$t_{DV}$ (TDVWH)	Data Valid to End of Write	15	—	25/25/20	—	30/30/25	—	35/35/30	—	40	—	ns
$t_{DH}$ (TWHDX)	Data Hold Time	3	—	5/5/3	—	5/5/3	—	5/5/3	—	5	—	ns
$t_{WZ}$ (TWLQZ)	Write Enable to Output in High Z	—	20	—	25	—	30	—	40	—	50	ns
$t_{OW}$ (TWHQX)	Output Active from End of Write	10	40	0/0/10	40/40/50	0/0/10	50/50/60	0/0/10	60/60/70	0	70	ns

**AC CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$  /  $-40^\circ C$  to  $+85^\circ C$  /  $0^\circ C$  to  $+70^\circ C$  unless otherwise noted.)<sup>(1)</sup>

SYMBOL	PARAMETER	IDT7168145 <sup>(2)</sup> IDT7168245		IDT7168155 IDT7168255		IDT7168170 IDT7168270		IDT7168185 IDT7168285		IDT71681100 <sup>(3)</sup> IDT71682100		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>												
$t_{RC}$ (TAVAV)	Read Cycle Time	45	—	55	—	70	—	85	—	100	—	ns
$t_{AA}$ (TAVQV)	Address Access Time	—	45	—	55	—	70	—	85	—	100	ns
$t_{ACS}$ (TELOV)	Chip Select Access Time	—	45	—	55	—	70	—	85	—	100	ns
$t_{OH}$ (TAXQX)	Output Hold from Address Change	5	—	5	—	5	—	5	—	5	—	ns
$t_{LZ}$ (TELQX)	Chip Selection to Output in Low Z	20	—	20	—	20	—	20	—	20	—	ns
$t_{HZ}$ (TEHQZ)	Chip Deselection to Output in High Z	—	20	—	25	—	30	—	40	—	50	ns
$t_{PU}$ (TELICCH)	Chip Selection to Power Up Time	0	—	0	—	0	—	0	—	0	—	ns
$t_{PD}$ (TEHICCL)	Chip Deselection to Power Down Time	—	40	—	55/55/50	—	70/70/60	—	85/85/70	—	100	ns
$t_{RCS}$	Read Command Set-Up Time	-5	—	-5	—	-5	—	-5	—	-5	—	ns
$t_{RCH}$	Read Command Hold Time	-5	—	-5	—	-5	—	-5	—	-5	—	ns
<b>WRITE CYCLE</b>												
$t_{WC}$ (TAVAV)	Write Cycle Time	40	—	45/45/50	—	55/55/60	—	65/65/70	—	80	—	ns
$t_{CW}$ (TELWH)	Chip Selection to End of Write	35	—	45	—	55	—	65	—	80	—	ns
$t_{AW}$ (TAVWH)	Address Valid to End of Write	35	—	45	—	55	—	65	—	80	—	ns
$t_{AS}$ (TAVWL)	Address Setup Time	0	—	0	—	0	—	0	—	0	—	ns
$t_{WP}$ (TWLWH)	Write Pulse Width	30	—	35	—	40	—	45	—	55	—	ns
$t_{WR}$ (TWHAX)	Write Recovery Time	0	—	0	—	0	—	0	—	0	—	ns
$t_{DV}$ (TDVWH)	Data Valid to End of Write	15	—	25/25/20	—	30/30/25	—	35/35/30	—	40	—	ns
$t_{DH}$ (TWHDX)	Data Hold Time	3	—	5/5/3	—	5/5/3	—	5/5/3	—	5	—	ns
$t_{LV}$ (TDVQV)	Data Valid to Output Valid (71681 only)	—	35	—	35/35/40	—	40/40/45	—	45/45/50	—	50	ns
$t_{WV}$ (TWLQV)	Write Enable to Output Valid (71681 only)	—	35	—	35/35/40	—	40/40/45	—	45/45/50	—	50	ns
$t_{WZ}$ (TWLQZ)	Write Enable to Output in High Z (71682 only)	—	—	—	25	—	30	—	40	—	50	ns
$t_{OW}$ (TWHQX)	Output Active from End of Write (71682 only)	—	—	0	40	0	50	0	60	0	70	ns

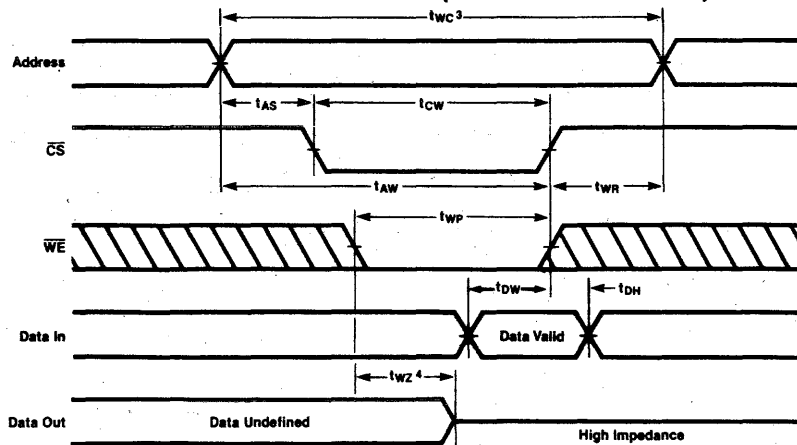
<sup>(1)</sup>Parameters listing three limits apply in this Temp. Range order: Military/Industrial/Commercial. All other limits apply to all three Temp. Ranges.

<sup>(2)</sup>Available in Commercial  $0^\circ C$  to  $+70^\circ C$  only.

<sup>(3)</sup>Available in Military  $-55^\circ C$  to  $+125^\circ C$  and Industrial  $-40^\circ C$  to  $+85^\circ C$  only.

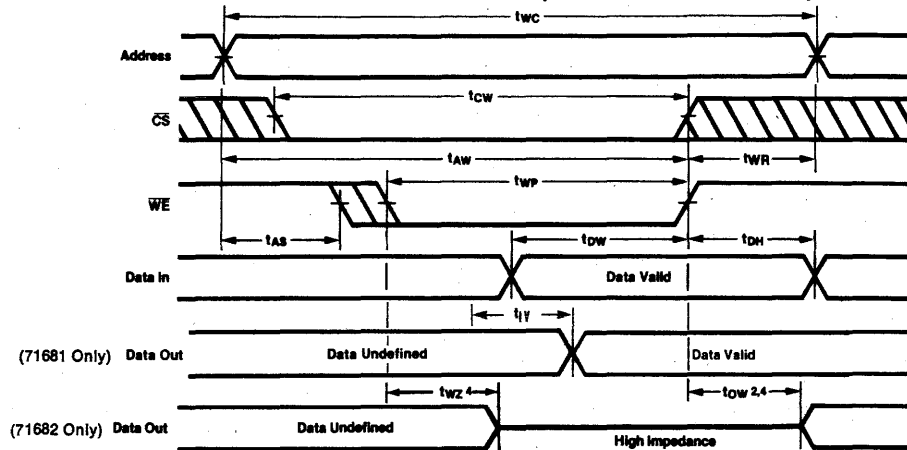


**TIMING WAVEFORM OF WRITE CYCLE NO. 2 ( $\overline{CS}$  CONTROLLED)<sup>1</sup> For IDT6168**

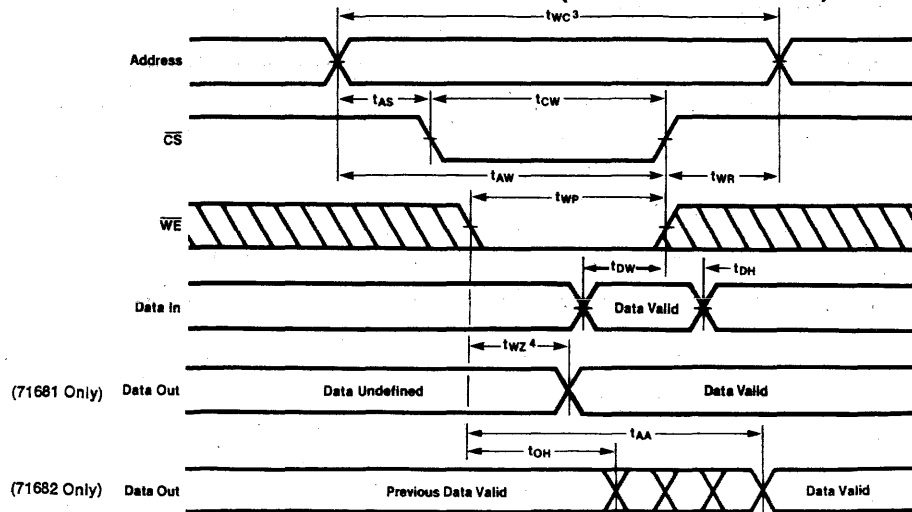


- NOTES: 1.  $\overline{CS}$  or  $\overline{WE}$  must be high during address transitions.  
 2. If  $\overline{CS}$  goes high simultaneously with  $\overline{WE}$  high, the output remains in a high impedance state.  
 3. All write cycle timings are referenced from the last valid address to the first transitioning address.  
 4. Transition is measured  $\pm 200\text{mV}$  from steady state voltage with specified loading in Figure 2. This parameter is sampled and not 100% tested.

**TIMING WAVEFORM OF WRITE CYCLE NO. 1 ( $\overline{WE}$  CONTROLLED)<sup>1</sup> For IDT71681 and IDT71682**

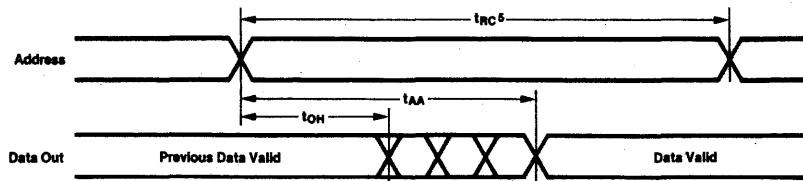


**TIMING WAVEFORM OF WRITE CYCLE NO. 2 ( $\overline{CS}$  CONTROLLED)<sup>1</sup> For IDT71681 and IDT71682**

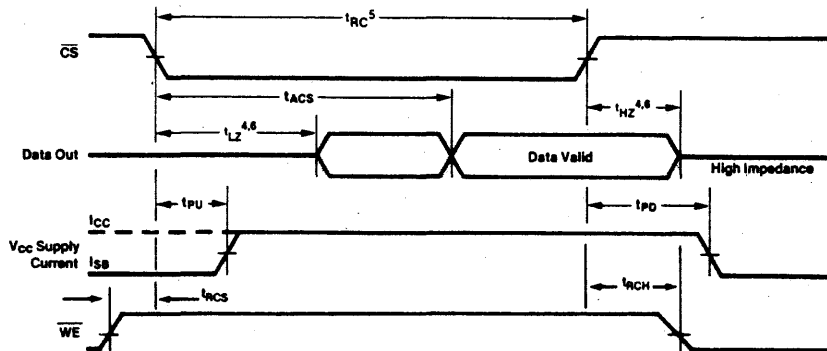


- NOTES: 1.  $\overline{CS}$  or  $\overline{WE}$  must be high during address transitions.  
 2. If  $\overline{CS}$  goes high simultaneously with  $\overline{WE}$  high, the output remains in a high impedance state.  
 3. All write cycle timings are referenced from the last valid address to the first transitioning address.  
 4. Transition is measured  $\pm 200\text{mV}$  from steady state voltage with specified loading in Figure 2. This parameter is sampled and not 100% tested.

**TIMING WAVEFORM OF READ CYCLE NO. 1<sup>(1,2)</sup>** For IDT6168 and IDT71681 and IDT71682

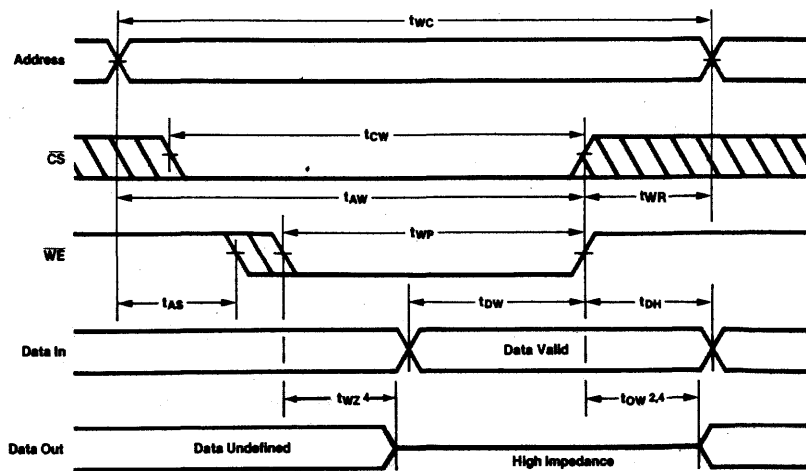


**TIMING WAVEFORM OF READ CYCLE NO. 2<sup>(1,3)</sup>** For IDT6168 and IDT71681 and IDT71682



- NOTES:
1.  $\overline{WE}$  is high for READ cycle.
  2.  $\overline{CS}$  is low for READ cycle.
  3. Address valid prior to or coincident with  $\overline{CS}$  transition low.
  4. Transition is measured  $\pm 500mV$  from steady state voltage with specified loading in Figure 2. This parameter is sampled, not 100% tested.
  5. All READ cycle timings are referenced from the last valid address to the first transitioning address.
  6. For any given speed grade, operating voltage and temperature,  $t_{HZ}$  will be less than or equal to  $t_{LZ}$ .

**TIMING WAVEFORM OF WRITE CYCLE NO. 1 ( $\overline{WE}$  CONTROLLED)<sup>1</sup>** For IDT6168

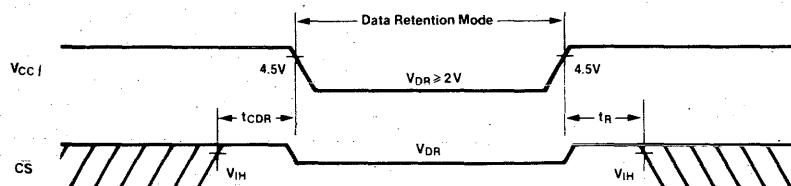


**LOW V<sub>CC</sub> DATA RETENTION CHARACTERISTICS** (T<sub>A</sub> = -55°C to +125°C/  
 -40°C to +85°C/0°C to +70°C)

SYMBOL	PARAMETER	TEST CONDITION		MIN.	TYP. <sup>1</sup>	MAX.	UNIT
V <sub>DR</sub>	V <sub>CC</sub> for Data Retention			2.0	—	—	V
I <sub>CCDR</sub>	Data Retention Current	$\overline{CS} \geq V_{CC} - 0.2V$ $V_{IN} \geq V_{CC} - 0.2V$ or $\leq 0.2V$	COM'L	—	0.5 <sup>(2)</sup> 1.0 <sup>(3)</sup>	20 <sup>(2)</sup> 30 <sup>(2)</sup>	μA
			IND	—	0.5 <sup>(2)</sup> 1.0 <sup>(3)</sup>	40 <sup>(2)</sup> 60 <sup>(3)</sup>	
			MIL	—	0.5 <sup>(2)</sup> 1.0 <sup>(3)</sup>	300 <sup>(2)</sup> 450 <sup>(3)</sup>	
t <sub>CDR</sub>	Chip Deselect to Data Retention Time			0	—	—	ns
t <sub>R</sub>	Operation Recovery Time			t <sub>RC</sub> <sup>(4)</sup>	—	—	ns

NOTES: 1. T<sub>A</sub> = 25°C      3. at V<sub>CC</sub> = 3V  
 2. at V<sub>CC</sub> = 2V      4. t<sub>RC</sub> = Read Cycle Time

**LOW V<sub>CC</sub> DATA RETENTION WAVEFORM**





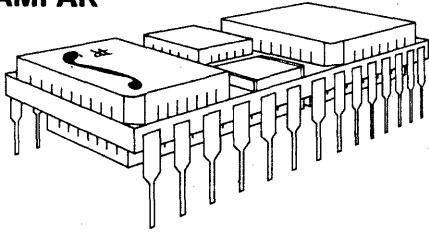
Integrated Device Technology, Inc.

# 64K (8K x 8) CMOS STATIC RAMPAK

# IDT7M864 IDT8M864

## MILITARY / COMMERCIAL TEMPERATURE RANGES

### 64K RAMPAK



#### FEATURES:

- 8,192 x 8 static RAM module complete with decoder and decoupling capacitor
- High-speed 85 (commercial only)/120/150/200ns (equal access and cycle times)
- Low power consumption; 300 mW typ.
- Two pinout options (64K EPROM & 64K static RAM)
- Utilizes IDT6116s — high performance 16K RAMs produced with advanced CEMOS™ technology
- CEMOS I process virtually eliminates alpha particle soft error rates (with no organic die coating)
- Single 5V ( $\pm 10\%$ ) power supply
- Input and output directly TTL-compatible
- Static operation: no clock or refresh required
- Standard military components 100% screened to MIL-STD-883 Class C
- Military components available 100% screened to Class B

#### DESCRIPTION:

The IDT7M864/IDT8M864 are 64K (8,192 x 8 bit) high speed static RAMs constructed on a ceramic substrate using 4 IDT6116 (2,048 x 8) static RAMs in leadless chip carriers. Functional equivalence to a monolithic 64K static

RAM is achieved by utilization of an on-board decoder circuit that interprets the higher order addresses  $A_{11}$  and  $A_{12}$  to select one of the four 2Kx8 RAMs. Extremely fast speeds can be achieved with this technique due to use of the IDT6116 fabricated in IDT's high performance, high-reliability technology - CEMOS I. This state-of-the-art technology, combined with innovative circuit design techniques, provides the fastest 16K static RAMs available.

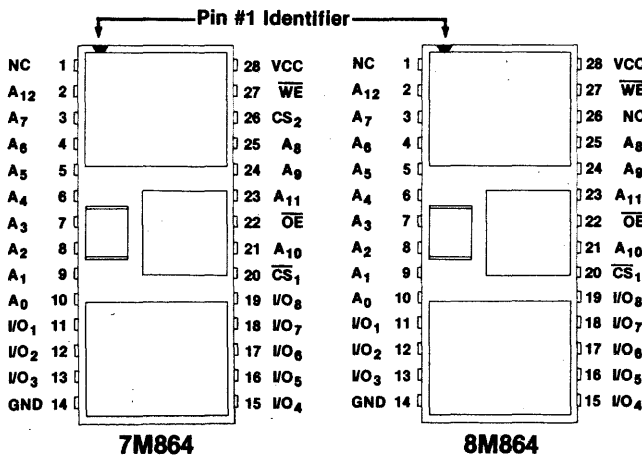
The IDT7M864/IDT8M864 are available with access times as fast as 85ns for commercial and 120ns for military temperature ranges, with maximum power consumption of only 825mW. The circuit also offers a reduced power standby mode. When  $\overline{CS}_1$  goes high, and/or  $\overline{CS}_2$  (7M864) goes low, the circuit will automatically go to, and remain in, a standby mode as long as these conditions are held. In the standby mode, the module consumes less than 275mW. Substantially lower power levels can be achieved in the ISB<sub>1</sub> mode (less than 25mW max.) and 2V data retention mode (less than 3mW max.) - see "DC Characteristics" and "Data Retention Characteristics" for details.

Pinout of the IDT8M864 is equivalent to the 64K EPROMs (no connect on pin 26), ideal for applications requiring easy microcode changes during prototyping. The IDT7M864's pinout is compatible with monolithic 64K static RAMs ( $\overline{CS}_2$  on pin 26).

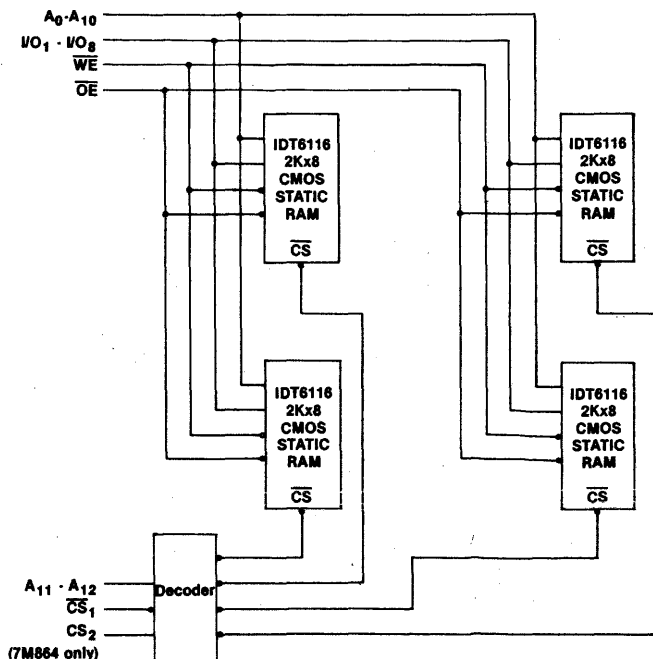
All inputs and outputs of the IDT7M864/IDT8M864 are TTL-compatible and operate from a single 5V supply, thus simplifying system designs. Full asynchronous circuitry is used, requiring no clocks or refreshing for operation, and provides equal access and cycle times for ease of use.

All IDT military module components are 100% processed to the test methods of MIL-STD-883, Method 5004, making them ideally suited to applications demanding the highest level of performance and reliability.

### PIN CONFIGURATIONS



### FUNCTIONAL BLOCK DIAGRAM



#### PIN NAMES

$A_0-A_{12}$	ADDRESS	$\overline{WE}$	WRITE ENABLE
$I/O_1-I/O_8$	DATA INPUT/OUTPUT	$\overline{OE}$	OUTPUT ENABLE
$\overline{CS}_1, \overline{CS}_2$	CHIP SELECT	GND	GROUND
$V_{CC}$	POWER		

CEMOS is a trademark of Integrated Device Technology, Inc.

Integrated Device Technology

MEMORY

**TRUTH TABLE**

MODE	$\overline{CS}_1$	$CS_2$	$\overline{OE}$	$\overline{WE}$	I/O OPERATION
Standby	H	X	X	X	High Z
Standby	X	L	X	X	High Z
Read	L	H	L	H	$D_{OUT}$
Read	L	H	H	H	High Z
Write	L	H	X	L	$D_{IN}$

**RECOMMENDED DC OPERATING CONDITIONS**

( $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  and  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ )

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
$V_{CC}$	Supply Voltage	4.5	5.0	5.5	V
GND	Supply Voltage	0	0	0	V
$V_{IH}$	Input High Voltage	2.2	3.5	6.0	V
$V_{IL}$	Input Low Voltage	-0.5*	—	.65	V
$C_L$	Output Load	—	—	100	pF
TTL	Output Load	—	—	1	—

\*  $V_{IL}$  min. = -3.5V for pulse widths less than 30 ns and duty cycles less than 50%.

**DC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  and  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ )

SYMBOL	PARAMETER	TEST CONDITIONS	IDT7M864/8M864			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	
$I_{L1}$	Input Leakage Current	$V_{CC} = 5.5V, V_{IN} = 0V$ to $V_{CC}$	—	—	15	$\mu\text{A}$
$I_{L0}$	Output Leakage Current	$\overline{OE}$ or $\overline{CS}_1 \geq V_{IH}$ , or $CS_2 \leq V_{IL}, V_{OUT} = 0V$ to $V_{CC}$	—	—	15	$\mu\text{A}$
$I_{CC}$	Operating Power Supply Current	$\overline{CS}_1 \leq V_{IL}, CS_2 \geq V_{IH}$ , Output Open	—	65	150	mA
$I_{CC1}$	Dynamic Operating Current	Min. Duty Cycle = 100%	—	65	150	mA
$I_{SB}$	Standby Power Supply Current	$\overline{CS}_1 \geq V_{IH}$ , or $CS_2 \leq V_{IL}$	—	20	50	mA
$I_{SB1}$	Full Standby Power Supply Current	$\overline{CS}_1 \geq V_{CC} - 0.2V$ , and/or $CS_2 \leq 0.2V$ $V_{IN} \geq V_{CC} - 0.2V$ or $\leq 0.2V$	—	15	1200	$\mu\text{A}$
$V_{OL}$	Output Low Voltage	$I_{OL} = 4mA$	—	—	0.4	V
$V_{OH}$	Output High Voltage	$I_{OH} = -1mA$	2.4	—	—	V

1.  $V_{CC} = 5V, T_A = 25^\circ\text{C}$

**ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

SYMBOL	RATING	COMMERCIAL	MILITARY	UNIT
$V_{TERM}$	Terminal Voltage with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	V
$T_A$	Operating Temperature	0 to +70	-55 to +125	$^\circ\text{C}$
$T_{BIAS}$	Temperature Under Bias	-10 to +85	-65 to +135	$^\circ\text{C}$
$T_{STG}$	Storage Temperature	-55 to +125	-65 to +150	$^\circ\text{C}$
$P_T$	Power Dissipation	4.0	4.0	W
$I_{OUT}$	DC Output Current	50	50	mA

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

**AC TEST CONDITIONS**

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	10ns
Input and Output Timing Reference Levels	1.5V
Output Load	1 TTL Gate and $C_L = 100\text{pF}$ (including scope and jig)

**CAPACITANCE** ( $T_A = 25^\circ\text{C}, f = 1.0\text{MHz}$ )

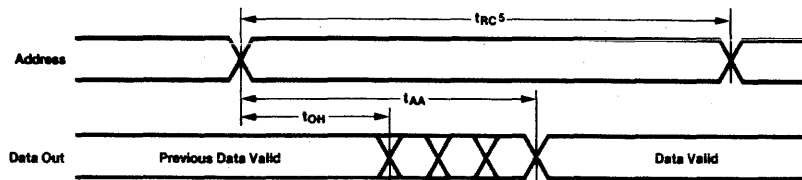
SYMBOL	ITEM	CONDITIONS	MAX.	UNIT
$C_{IN}$	Input Capacitance	$V_{IN} = 0V$	20	pF
$C_{OUT}$	Output Capacitance	$V_{OUT} = 0V$	22	pF

NOTE: This parameter is sampled and not 100% tested.

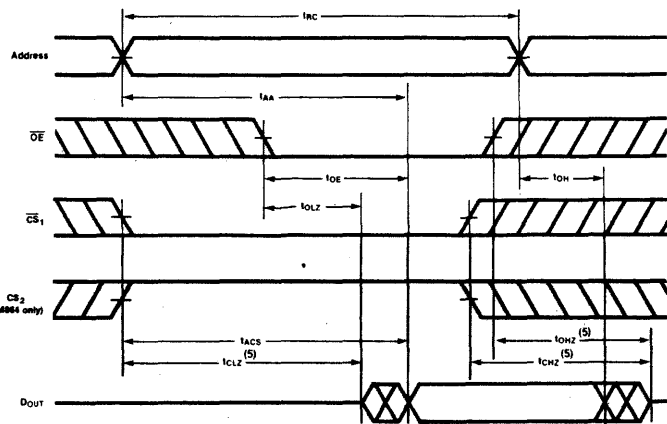
**AC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$  and  $0^\circ C$  to  $+70^\circ C$ )

SYMBOL	PARAMETER	7M/8M864L85 COMMERCIAL ONLY		7M/8M864L120		7M/8M864L150		7M/8M864L200		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>										
$t_{RC}$	Read Cycle Time	85	—	120	—	150	—	200	—	ns
$t_{AA}$	Address Access Time	—	85	—	120	—	150	—	200	ns
$t_{ACS}$	Chip Select Access Time	—	85	—	120	—	150	—	200	ns
$t_{CLZ}$	Chip Selection to Output in Low Z	5	—	5	—	5	—	5	—	ns
$t_{OE}$	Output Enable to Output Valid	—	65	—	65	—	80	—	100	ns
$t_{OLZ}$	Output Enable to Output in Low Z	0	—	0	—	0	—	0	—	ns
$t_{CHZ}$	Chip Deselection to Output in High Z	—	55	—	70	—	70	—	80	ns
$t_{OHZ}$	Output Disable to Output in High Z	—	40	—	40	—	40	—	50	ns
$t_{OH}$	Output Hold from Address Change	0	—	0	—	0	—	0	—	ns
<b>WRITE CYCLE</b>										
$t_{WC}$	Write Cycle Time	85	—	120	—	150	—	200	—	ns
$t_{CW}$	Chip Selection to End of Write	65	—	80	—	100	—	120	—	ns
$t_{AW}$	Address Valid to End of Write	70	—	85	—	100	—	120	—	ns
$t_{AS}$	Address Setup Time	15	—	15	—	20	—	20	—	ns
$t_{WP}$	Write Pulse Width	55	—	55	—	70	—	90	—	ns
$t_{WR}$	Write Recovery Time	10	—	10	—	10	—	10	—	ns
$t_{OHZ}$	Output Disable to Output in High Z	—	40	—	40	—	40	—	50	ns
$t_{WHZ}$	Write to Output in High Z	0	50	0	50	0	60	0	60	ns
$t_{DW}$	Data to Write Time Overlap	30	—	30	—	35	—	40	—	ns
$t_{DH}$	Data Hold from Write Time	10	—	10	—	10	—	10	—	ns
$t_{OW}$	Output Active from End of Write	0	—	0	—	0	—	5	—	ns

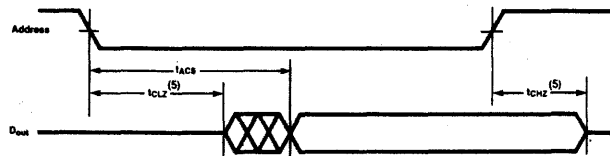
**TIMING WAVEFORM OF READ CYCLE NO. 1** <sup>(1,2)</sup>



**TIMING WAVEFORM OF READ CYCLE NO. 2** <sup>(1,3)</sup>



**TIMING WAVEFORM OF READ CYCLE NO. 3** <sup>(1,3,4)</sup>

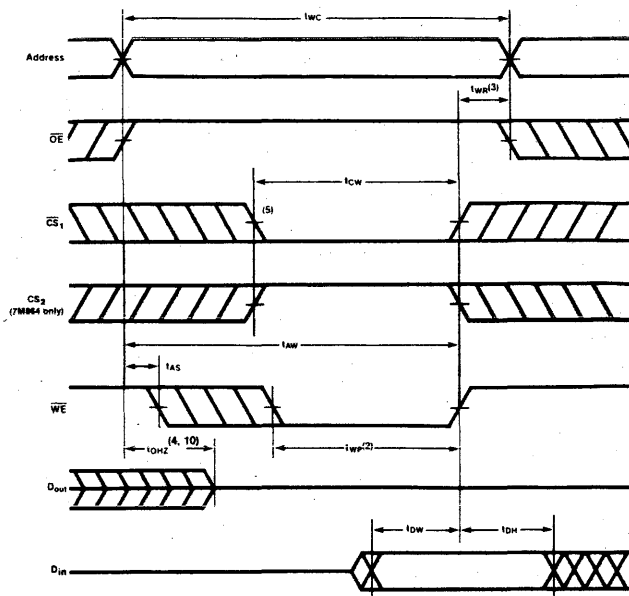


- NOTES: 1.  $\overline{WE}$  is High for Read Cycle.  
 2. Device is continuously selected,  $\overline{CS}_1 = V_{IL}$ ,  $CS_2 = V_{IH}$  (7M864 only).  
 3. Address valid prior to or coincident with  $\overline{CS}_1$  transition low,  $CS_2$  transition high (7M864 only).  
 4.  $\overline{OE} = V_{IL}$ .  
 5. Transition is measured  $\pm 500mV$  from steady state. This parameter is sampled and not 100% tested.

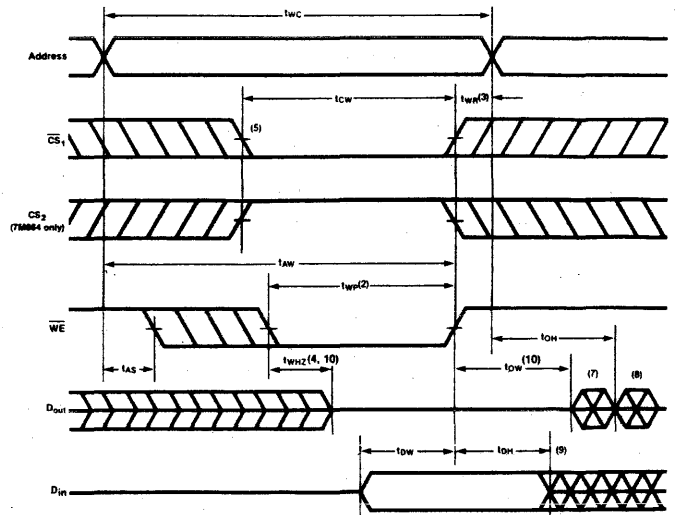
Integrated Device Technology

MEMORY

**TIMING WAVEFORM OF WRITE CYCLE 1<sup>(1)</sup>**



**TIMING WAVEFORM OF WRITE CYCLE 2<sup>(1,6)</sup>**



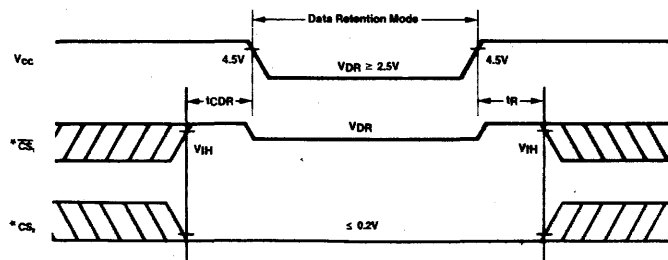
- NOTES: 1.  $\overline{WE}$  is high during all address transitions.  
 2. A write occurs during the overlap ( $t_{WP}$ ) of a low  $\overline{CS}_1$ , or high  $CS_2$  (7M864 only) and a low  $\overline{WE}$ .  
 3.  $t_{WR}$  is measured from the earlier of  $\overline{CS}_1$  or  $\overline{WE}$  going high or  $CS_2$  going low to the end of write cycle.  
 4. During this period, I/O pins are in the output state so that the input signals of opposite phase to the outputs must not be applied.  
 5. If the  $\overline{CS}_1$  low transition or  $CS_2$  high transition occurs simultaneously with the  $\overline{WE}$  low transitions or after the  $\overline{WE}$  transition, outputs remain in a high impedance state.  
 6.  $\overline{OE}$  is continuously low ( $\overline{OE} = V_{IL}$ ).  
 7.  $D_{OUT}$  is the same phase of write data of this write cycle.  
 8.  $D_{OUT}$  is the read data of next address.  
 9. If  $\overline{CS}_1$  is low or  $CS_2$  is high during this period, I/O pins are in the output state. Then the data input signals of opposite phase to the outputs must not be applied to them.  
 10. Transition is measured  $\pm 500mV$  from steady state. This parameter is sampled and not 100% tested.

**LOW  $V_{CC}$  DATA RETENTION CHARACTERISTICS ( $T_A = -55^\circ C$  to  $+125^\circ C$  and  $0^\circ C$  to  $+70^\circ C$ )**

SYMBOL	PARAMETER	TEST CONDITION	MIN.	TYP. <sup>(1)</sup>	MAX. COMM.	MAX. MIL.	UNIT
$V_{DR}$	$V_{CC}$ for Data Retention		2.5	—	—	—	V
$I_{CCDR}$	Data Retention Current	$\overline{CS}_1 \geq V_{CC} - 0.2V, CS_2 \leq 0.2V$ $V_{IN} \geq V_{CC} - 0.2V$ or $\leq 0.2V$	—	2.0 <sup>(2)</sup>	80 <sup>(2)</sup>	850 <sup>(2)</sup>	$\mu A$
			—	4.0 <sup>(3)</sup>	120 <sup>(3)</sup>	1000 <sup>(3)</sup>	
$t_{CDR}$	Chip Deselect to Data Retention Time		0	—	—	—	ns
$t_R$	Operation Recovery Time		$t_{RC}$ <sup>(4)</sup>	—	—	—	ns

- NOTES: 1.  $T_A = 25^\circ C$  3. at  $V_{CC} = 3V$   
 2. at  $V_{CC} = 2.5V$  4.  $t_{RC}$  = Read Cycle Time

**LOW  $V_{CC}$  DATA RETENTION WAVEFORM**



\*Low  $V_{CC}$  data retention achieved by the indicated  $\overline{CS}_1$  waveform or  $CS_2$  waveform.



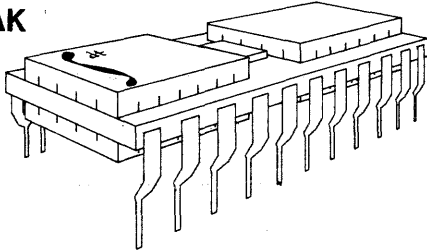
Integrated Device Technology, Inc.

# 64K (16K x 4) CMOS STATIC RAMPAK

# IDT7M464

## MILITARY / COMMERCIAL TEMPERATURE RANGES

### 64K RAMPAK



#### FEATURES:

- 65,536x1 bit static RAM module complete with decoder and decoupling capacitor
- High-speed 70 (commercial only)/100/120/150ns (equal access and cycle times)
- Low power consumption; 250mW typically
- Offered in a 22 pin, 300 mil center sidebrazed DIP
- Pin compatible with proposed monolithics
- Utilizes IDT6167s — high performance 16K RAMs produced with advanced CEMOS™ technology
- CEMOS I process virtually eliminates alpha particle soft error rates (with no organic die coating)
- Single 5V (± 10%) power supply
- Input and output directly TTL-compatible
- Static operation: no clock or refresh required
- Standard military components 100% screened to MIL-STD-883 Class C
- Military components available 100% screened to Class B

#### DESCRIPTION:

The IDT7M464 is a 64K (16,384 x 4 bit) high speed CMOS static RAM constructed on a ceramic substrate using 4

IDT6167 (16,384 x 1 bit) CMOS static RAMs in leadless chip carriers. Functional equivalence to proposed monolithic 16Kx4 static RAMs is achieved by utilization of tungsten traces within the substrate to connect the four 16K RAMs in a 16Kx4 configuration. Extremely fast speeds can be achieved with this technique due to use of the IDT6167 fabricated in IDT's high-performance, high-reliability technology — CEMOS™. This state-of-the-art technology, combined with innovative circuit design techniques, provides the fastest 16K static RAMs available.

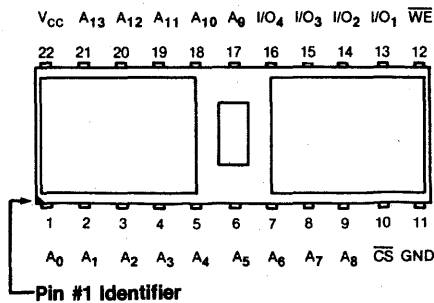
The IDT7M464 is available with access times as fast as 55ns for commercial and 65ns for military temperature ranges, with maximum power consumption of only 1.1 watt. The circuit also offers a reduced power standby mode. When  $\overline{CS}$  goes high, the circuit will automatically go to, and remain in, a standby mode as long as  $\overline{CS}$  remains high, consuming only 300mW maximum. Substantially lower power levels can be achieved in the  $I_{SB1}$  mode (less than 6mW max.) and 2V data retention mode (less than 2mW max.) — see "DC Characteristics" and "Data Retention Characteristics" for details.

The IDT7M464 is offered in a space saving 22-pin, 300 mil pin center package, providing equivalent pinout to the proposed monolithic 16Kx4 static RAMs.

All inputs and outputs of the IDT7M464 are TTL-compatible and operate from a single 5V supply, thus simplifying system designs. Full asynchronous circuitry is used, requiring no clocks or refreshing for operation, and provides equal access and cycle times for ease of use.

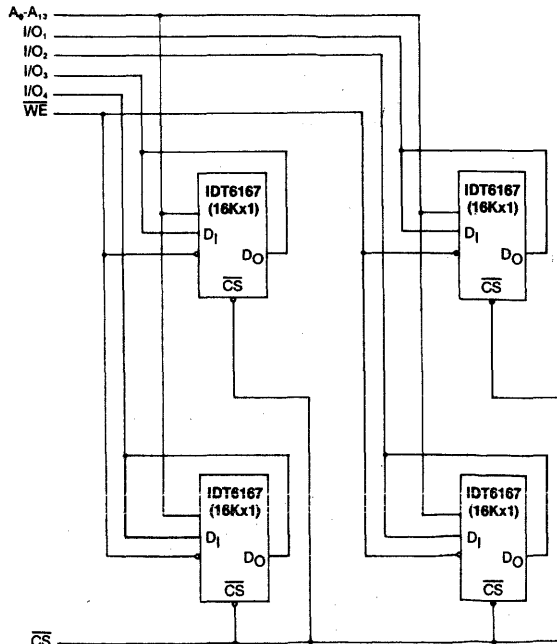
All IDT military module components are 100% processed to the test methods of MIL-STD-883, Method 5004, making them ideally suited to applications demanding the highest level of performance and reliability.

### PIN CONFIGURATION



7M464  
TOP VIEW

### FUNCTIONAL BLOCK DIAGRAM



#### PIN NAMES

$A_0-A_{13}$	ADDRESS	$\overline{WE}$	WRITE ENABLE
$I/O_1-I/O_4$	DATA INPUT/OUTPUT	$V_{CC}$	POWER
$\overline{CS}$	CHIP SELECT	GND	GROUND

CEMOS is a trademark of Integrated Device Technology, Inc.



**RECOMMENDED DC OPERATING CONDITIONS**

( $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  and  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ )

**TRUTH TABLE**

MODE	$\overline{\text{CS}}$	$\overline{\text{WE}}$	OUTPUT	POWER
Standby	H	X	High Z	Standby
Read	L	H	D Out	Active
Write	L	L	High Z	Active

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
$V_{CC}$	Supply Voltage	4.5	5.0	5.5	V
GND	Supply Voltage	0	0	0	V
$V_{IH}$	Input High Voltage	2.2	3.5	6.0	V
$V_{IL}$	Input Low Voltage	-0.5*	—	+0.8	V

\*  $V_{IL}$  min. = -3.5V for pulse widths less than 30 ns and duty cycles less than 50%.

**DC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  and  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ )

SYMBOL	PARAMETER	TEST CONDITIONS	IDT7M164			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	
$I_{L1}$	Input Leakage Current	$V_{CC} = 5.5V$ , $V_{IN} = 0V$ to $V_{CC}$	—	—	15	$\mu\text{A}$
$I_{L0}$	Output Leakage Current	$\overline{\text{CS}} = V_{IH}$ , $V_{OUT} = 0V$ to $V_{CC}$	—	—	15	$\mu\text{A}$
$I_{CC1}$	Operating Power Supply Current	$\overline{\text{CS}} = V_{IL}$ , Output Open	—	100	195	mA
$I_{CC2}$	Dynamic Operating Current	Min. Duty Cycle = 100%	—	100	195	mA
$I_{SB}$	Standby Power Supply Current	$\overline{\text{CS}} \geq V_{IH}$	—	20	55	mA
$I_{SB1}$	Full Standby Power Supply Current	$\overline{\text{CS}} \geq V_{CC} - 0.2V$ $V_{IN} \geq V_{CC} - 0.2V$ or $\leq 0.2V$	—	8	1000 <sup>(2)</sup>	$\mu\text{A}$
$V_{OL}$	Output Low Voltage	$I_{OL} = 8\text{mA}$	—	—	0.4	V
$V_{OH}$	Output High Voltage	$I_{OH} = -4\text{mA}$	2.4	—	—	V

- $V_{CC} = 5V$ ,  $T_A = 25^\circ\text{C}$
- $I_{SB1}$  max at commercial temperature = 0.2mA

**ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

SYMBOL	RATING	COMMERCIAL	MILITARY	UNIT
$V_{TERM}$	Terminal Voltage with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	V
$T_A$	Operating Temperature	0 to +70	-55 to +125	$^\circ\text{C}$
$T_{BIAS}$	Temperature Under Bias	-10 to +85	-65 to +135	$^\circ\text{C}$
$T_{STG}$	Storage Temperature	-55 to +125	-65 to +150	$^\circ\text{C}$
$P_T$	Power Dissipation	4.0	4.0	W
$I_{OUT}$	DC Output Current	50	50	mA

- Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

**AC TEST CONDITIONS**

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	5ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	See Figures 1 and 2

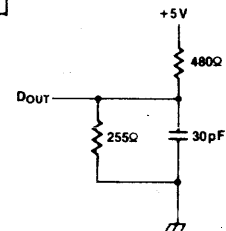


Figure 1. Output Load

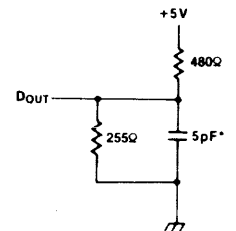


Figure 2. Output Load (for  $t_{HZ}$ ,  $t_{LZ}$ ,  $t_{WZ}$ , and  $t_{OW}$ )

**CAPACITANCE** ( $T_A = 25^\circ\text{C}$ ,  $f = 1.0\text{MHz}$ )

SYMBOL	ITEM	CONDITIONS	MAX.	UNIT
$C_{IN}$	Input Capacitance	$V_{IN} = 0V$	20	pF
$C_{OUT}$	Output Capacitance	$V_{OUT} = 0V$	22	pF

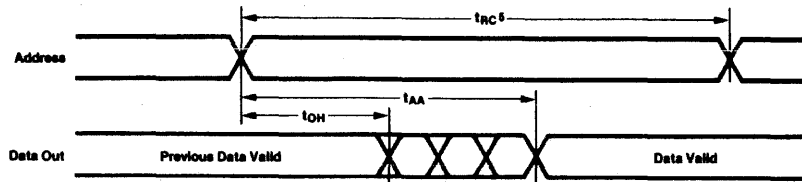
NOTE: This parameter is sampled and not 100% tested.

\*Including scope and jig.

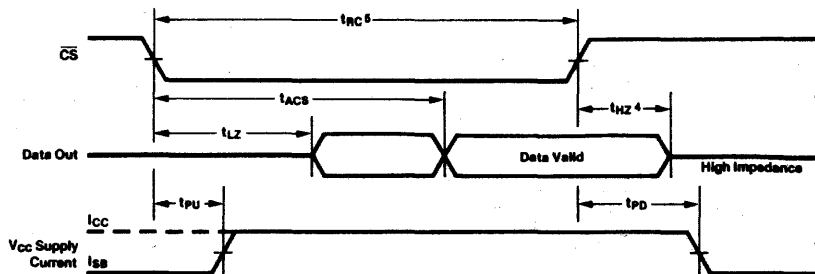
**AC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$  and  $0^\circ C$  to  $+70^\circ C$ )

SYMBOL	PARAMETER	7M464S55 COMMERCIAL ONLY		7M464S65		7M464S85		7M464S100 MILITARY ONLY		UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>										
$t_{RC}$	Read Cycle Time	55	—	65	—	85	—	100	—	ns
$t_{AA}$	Address Access Time	—	55	—	65	—	85	—	100	ns
$t_{ACS}$	Chip Select Access Time	—	55	—	65	—	85	—	100	ns
$t_{OH}$	Output Hold from Address Change	5	—	5	—	5	—	5	—	ns
$t_{LZ}$	Chip Selection to Output in Low Z	5	—	5	—	5	—	5	—	ns
$t_{HZ}$	Chip Deselection to Output in High Z	0	40	0	40	0	50	0	50	ns
$t_{PU}$	Chip Selection to Power Up Time	0	—	0	—	0	—	0	—	ns
$t_{PD}$	Chip Deselection to Power Down Time	—	55	—	65	—	85	—	100	ns
<b>WRITE CYCLE</b>										
$t_{WC}$	Write Cycle Time	55	—	65	—	85	—	100	—	ns
$t_{CW}$	Chip Selection to End of Write	45	—	55	—	65	—	80	—	ns
$t_{AW}$	Address Valid to End of Write	45	—	55	—	65	—	80	—	ns
$t_{AS}$	Address Setup Time	0	—	0	—	0	—	0	—	ns
$t_{WP}$	Write Pulse Width	35	—	40	—	45	—	55	—	ns
$t_{WR}$	Write Recovery Time	0	—	0	—	0	—	0	—	ns
$t_{DW}$	Data Valid to End of Write	25	—	30	—	35	—	40	—	ns
$t_{DH}$	Data Hold Time	0	—	0	—	0	—	0	—	ns
$t_{WZ}$	Write Enable to Output in High Z	0	40	0	40	0	50	0	50	ns
$t_{OW}$	Output Active from End of Write	0	—	0	—	0	—	0	—	ns

**TIMING WAVEFORM OF READ CYCLE NO. 1** <sup>(1,2)</sup>

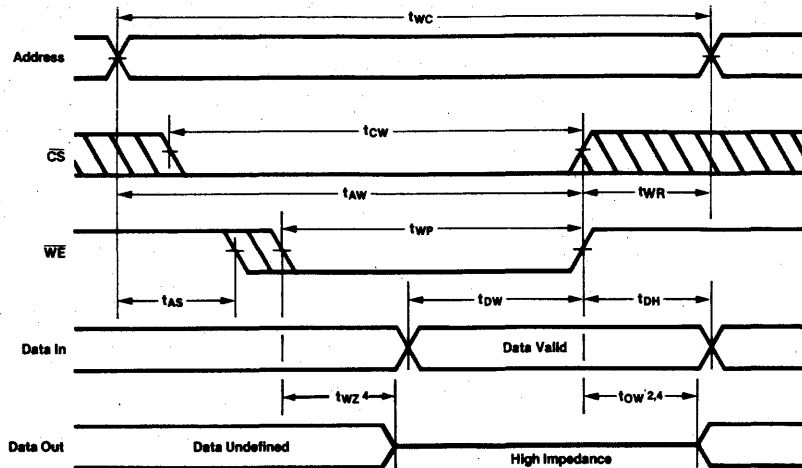


**TIMING WAVEFORM OF READ CYCLE NO. 2** <sup>(1,3)</sup>

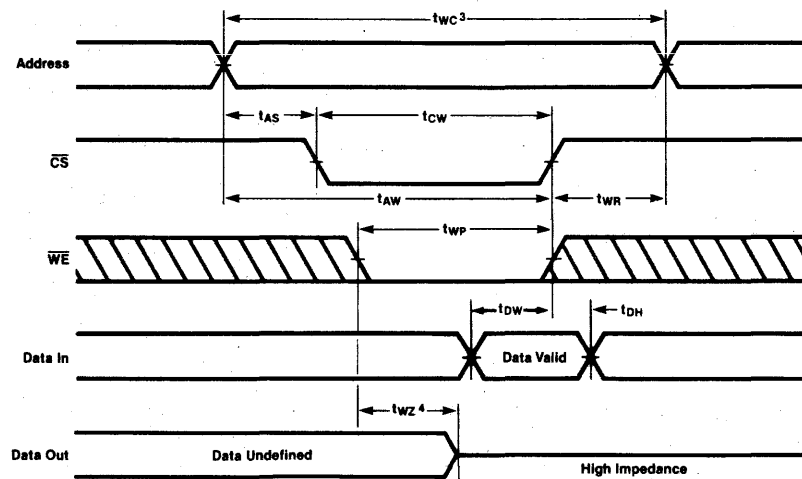


- NOTES:
- $\overline{WE}$  is high for READ cycle.
  - $\overline{CS}$  is low for READ cycle.
  - Address valid prior to or coincident with  $\overline{CS}$  transition low.
  - Transition is measured  $\pm 500mV$  from steady state voltage with specified loading in Figure 2. This parameter is sampled, not 100% tested.
  - All READ cycle timings are referenced from the last valid address to the first transitioning address.

**TIMING WAVEFORM OF WRITE CYCLE NO. 1 ( $\overline{WE}$  CONTROLLED)<sup>1</sup>**



**TIMING WAVEFORM OF WRITE CYCLE NO. 2 ( $\overline{CS}$  CONTROLLED)<sup>1</sup>**



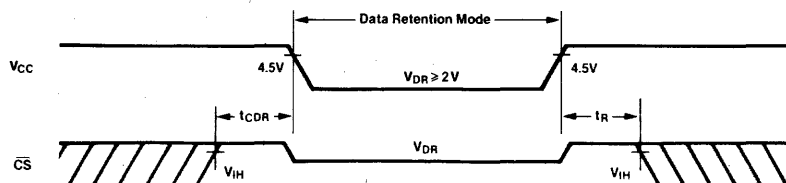
- NOTES: 1.  $\overline{CS}$  or  $\overline{WE}$  must be high during address transitions.  
 2. If  $\overline{CS}$  goes high simultaneously with  $\overline{WE}$  high, the output remains in a high impedance state.  
 3. All write cycle timings are referenced from the last valid address to the first transitioning address.  
 4. Transition is measured  $\pm 200mV$  from steady state voltage with specified loading in Figure 2. This parameter is sampled and not 100% tested.

**LOW  $V_{CC}$  DATA RETENTION CHARACTERISTICS ( $T_A = -55^\circ C$  to  $+125^\circ C$  and  $0^\circ C$  to  $+70^\circ C$ )**

SYMBOL	PARAMETER	TEST CONDITION	MIN.	TYP. <sup>(1)</sup>	MAX. COMM.	MAX. MIL.	UNIT
$V_{DR}$	$V_{CC}$ for Data Retention		2.0	—	—	—	V
$I_{CCDR}$	Data Retention Current	$\overline{CS} \geq V_{CC} - 0.2V$	—	2.0 <sup>(2)</sup> 4.0 <sup>(3)</sup>	80 <sup>(2)</sup> 120 <sup>(3)</sup>	850 <sup>(2)</sup> 1000 <sup>(3)</sup>	$\mu A$
$t_{CDR}$	Chip Deselect to Data Retention Time	$V_{IN} \geq V_{CC} - 0.2V$ or $\leq 0.2V$	15	—	—	—	ns
$t_R$	Operation Recovery Time		$t_{RC}$ <sup>(4)</sup>	—	—	—	ns

- NOTES: 1.  $T_A = 25^\circ C$       3. at  $V_{CC} = 3V$   
 2. at  $V_{CC} = 2V$       4.  $t_{RC} =$  Read Cycle Time

**LOW  $V_{CC}$  DATA RETENTION WAVEFORM**





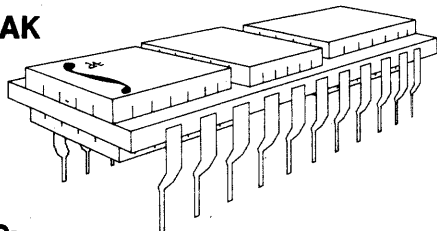
Integrated Device Technology, Inc.

# 64K (64K × 1) CMOS STATIC RAMPAK

# IDT7M164 PRELIMINARY

## MILITARY / COMMERCIAL TEMPERATURE RANGES

### 64K RAMPAK



#### FEATURES:

- 65,536 bit CMOS static RAM module with decoupling capacitor
- High-speed: 55/65/85ns max. commercial  
65/85/100ns max. military
- Low power consumption: 1.1W max.
- Pinout identical to proposed monolithic 64Kx1 static RAMs
- Utilizes IDT6167s — high performance 16K RAMs produced with advanced CEMOS™ I technology
- CEMOS I process virtually eliminates alpha particle soft error rates (with no organic die coating)
- Single 5V (± 10%) power supply
- Input and output directly TTL-compatible
- Static operation: no clock or refresh required
- Standard military components 100% screened to MIL-STD-883 Class C
- Military components available 100% screened to Class B

#### DESCRIPTION:

The IDT7M164 is a 64K (65,536x1 bit) high speed static RAM constructed on a ceramic substrate using 4 IDT6167 (16Kx1) static RAMs in leadless chip carriers. Functional

equivalence to proposed monolithic 64Kx1 static RAMs is achieved by utilization of an on-board decoder circuit that interprets the higher order addresses  $A_{14}$  and  $A_{15}$  to select one of the four 16Kx1 RAMs. Extremely fast speeds can be achieved with this technique due to use of the IDT6167 fabricated in IDT's high performance, high-reliability technology — CEMOS™ I. This state-of-the-art technology, combined with innovative circuit design techniques, provides the fastest 16K static RAMs available.

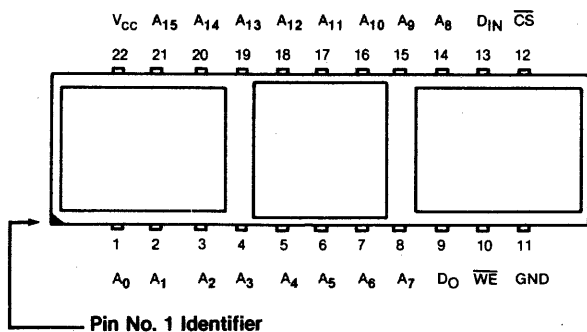
The IDT7M164 is available with access times as fast as 70ns for commercial and 100ns for military temperature ranges, with maximum power consumption of only 660mW. The circuit also offers a reduced power standby mode. When  $\overline{CS}$  goes high, the circuit will automatically go to, and remain in, a standby mode as long as this condition is held. In the standby mode, the module typically consumes less than 100 mW. Substantially lower power levels can be achieved in the ISB, mode (less than 50  $\mu$ W typ.) and 2V data retention mode (less than 4  $\mu$ W typ.) — see "DC Characteristics" and "Data Retention Characteristics" for details.

The IDT7M164 is offered in a space saving 22-pin, 300 mil pin center package, providing equivalent pinout to the proposed monolithic 64Kx1 static RAMs.

All inputs and outputs of the IDT7M164 are TTL-compatible and operate from a single 5V supply, thus simplifying system designs. Full asynchronous circuitry is used, requiring no clocks or refreshing for operation, and provides equal access and cycle times for ease of use.

All IDT military module components are 100% processed to the test methods of MIL-STD-883, Method 5004, making them ideally suited to applications demanding the highest level of performance and reliability.

### PIN CONFIGURATION

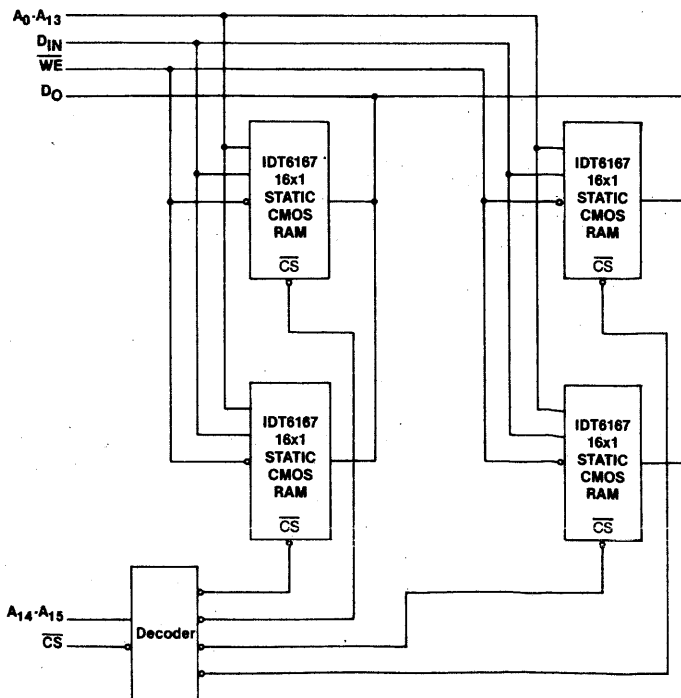


IDT7M164  
TOP VIEW

### PIN NAMES

$A_0$ - $A_{15}$	ADDRESS	$\overline{WE}$	WRITE ENABLE
$D_{IN}$	DATA INPUT	$D_0$	DATA OUTPUT
$\overline{CS}$	CHIP SELECT	GND	GROUND
$V_{CC}$	POWER		

### FUNCTIONAL BLOCK DIAGRAM



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## TRUTH TABLE

MODE	$\overline{CS}$	$\overline{WE}$	OUTPUT	POWER
Standby	H	X	High Z	Standby
Read	L	H	D Out	Active
Write	L	L	High Z	Active

## RECOMMENDED DC OPERATING CONDITIONS

 $(T_A = -55^\circ\text{C to } +125^\circ\text{C and } 0^\circ\text{C to } +70^\circ\text{C})$ 

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
$V_{CC}$	Supply Voltage	4.5	5.0	5.5	V
GND	Supply Voltage	0	0	0	V
$V_{IH}$	Input High Voltage	2.2	3.5	6.0	V
$V_{IL}$	Input Low Voltage	-0.5*	—	+0.8	V

\*  $V_{IL}$  min. = -3.5V for pulse widths less than 30 ns and duty cycles less than 50%.DC ELECTRICAL CHARACTERISTICS ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ\text{C to } +125^\circ\text{C and } 0^\circ\text{C to } +70^\circ\text{C}$ )

SYMBOL	PARAMETER	TEST CONDITIONS	IDT7M164			UNIT
			MIN.	TYP. <sup>(1)</sup>	MAX.	
$ I_{LI} $	Input Leakage Current	$V_{CC} = 5.5V, V_{IN} = 0V \text{ to } V_{CC}$	—	—	20	$\mu\text{A}$
$ I_{LO} $	Output Leakage Current	$\overline{CS} = V_{IH}, V_{OUT} = 0V \text{ to } V_{CC}$	—	—	20	$\mu\text{A}$
$I_{CC1}$	Operating Power Supply Current	$\overline{CS} = V_{IL}, \text{Output Open}$	—	50	120	mA
$I_{CC2}$	Dynamic Operating Current	Min. Duty Cycle = 100%	—	50	120	mA
$I_{SB}$	Standby Power Supply Current	$\overline{CS} \geq V_{IH}$	—	20	80	mA
$I_{SB1}$	Full Standby Power Supply Current	$\overline{CS} \geq V_{CC} - 0.2V$ $V_{IN} \geq V_{CC} - 0.2V \text{ or } \leq 0.2V$	—	0.008	3.5 <sup>(2)</sup>	mA
$V_{OL}$	Output Low Voltage	$I_{OL} = 8\text{mA}$	—	—	0.4	V
$V_{OH}$	Output High Voltage	$I_{OH} = -4\text{mA}$	2.4	—	—	V

1:  $V_{CC} = 5V, T_A = 25^\circ\text{C}$ 2:  $I_{SB1}$  max at commercial temperature = 0.2mAABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

SYMBOL	RATING	COMMERCIAL	MILITARY	UNIT
$V_{TERM}$	Terminal Voltage with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	V
$T_A$	Operating Temperature	0 to +70	-55 to +125	$^\circ\text{C}$
$T_{BIAS}$	Temperature Under Bias	-10 to +85	-65 to +135	$^\circ\text{C}$
$T_{STG}$	Storage Temperature	-55 to +125	-65 to +150	$^\circ\text{C}$
$P_T$	Power Dissipation	4.0	4.0	W
$I_{OUT}$	DC Output Current	50	50	mA

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

CAPACITANCE ( $T_A = 25^\circ\text{C}, f = 1.0\text{MHz}$ )

SYMBOL	ITEM	CONDITIONS	MAX.	UNIT
$C_{IN}$	Input Capacitance	$V_{IN} = 0V$	20	pF
$C_{OUT}$	Output Capacitance	$V_{OUT} = 0V$	22	pF

NOTE: This parameter is sampled and not 100% tested.

## AC TEST CONDITIONS

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	5ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	See Figures 1 and 2

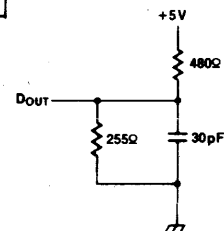
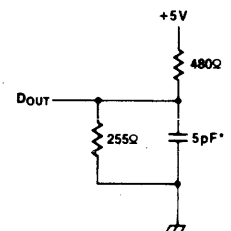


Figure 1. Output Load

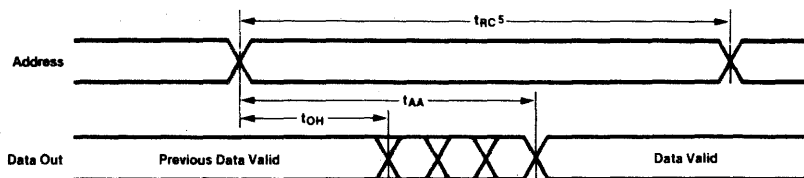
Figure 2. Output Load (for  $t_{H2}$ ,  $t_{L2}$ ,  $t_{W2}$ , and  $t_{OW}$ )

\*Including scope and jig.

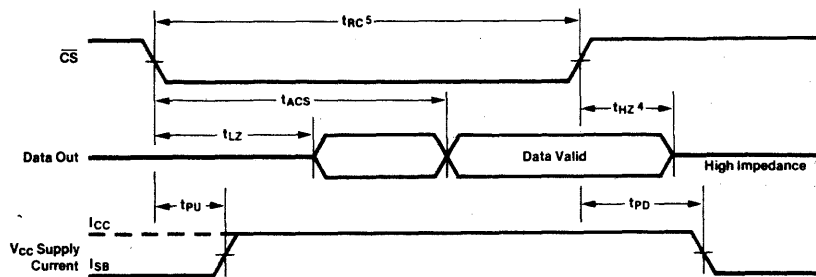
**AC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$  and  $0^\circ C$  to  $+70^\circ C$ )

SYMBOL	PARAMETER	COMMERCIAL ONLY								UNIT
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
<b>READ CYCLE</b>										
$t_{RC}$	Read Cycle Time	70	—	100	—	120	—	150	—	ns
$t_{AA}$	Address Access Time	—	70	—	100	—	120	—	150	ns
$t_{ACS}$	Chip Select Access Time	—	70	—	100	—	120	—	150	ns
$t_{OH}$	Output Hold from Address Change	10	—	10	—	10	—	10	—	ns
$t_{LZ}$	Chip Selection to Output in Low Z	20	—	35	—	40	—	55	—	ns
$t_{HZ}$	Chip Deselection to Output in High Z	15	55	30	70	35	80	50	100	ns
$t_{PU}$	Chip Selection to Power Up Time	15	—	30	—	35	—	50	—	ns
$t_{PD}$	Chip Deselection to Power Down Time	—	70	—	100	—	120	—	150	ns
<b>WRITE CYCLE</b>										
$t_{WC}$	Write Cycle Time	70	—	100	—	120	—	150	—	ns
$t_{CW}$	Chip Selection to End of Write	65	—	80	—	100	—	120	—	ns
$t_{AW}$	Address Valid to End of Write	65	—	80	—	100	—	120	—	ns
$t_{AS}$	Address Setup Time	0	—	0	—	0	—	0	—	ns
$t_{WP}$	Write Pulse Width	40	—	45	—	50	—	60	—	ns
$t_{WRW}$	Write Recovery Time $\overline{WE}$ Controlled	0	—	0	—	0	—	0	—	ns
$t_{WRC}$	Write Recovery Time $\overline{CS}$ Controlled	25	—	25	—	30	—	35	—	ns
$t_{DW}$	Data Valid to End of Write	30	—	30	—	35	—	40	—	ns
$t_{DHW}$	Data Hold Time $\overline{WE}$ Controlled	0	—	0	—	0	—	0	—	ns
$t_{DHC}$	Data Hold Time $\overline{CS}$ Controlled	25	—	25	—	30	—	35	—	ns
$t_{WZ}$	Write Enable to Output in High Z	0	40	0	40	0	50	0	50	ns
$t_{OW}$	Output Active from End of Write	0	—	0	—	0	—	0	—	ns

**TIMING WAVEFORM OF READ CYCLE NO. 1** <sup>(1,2)</sup>

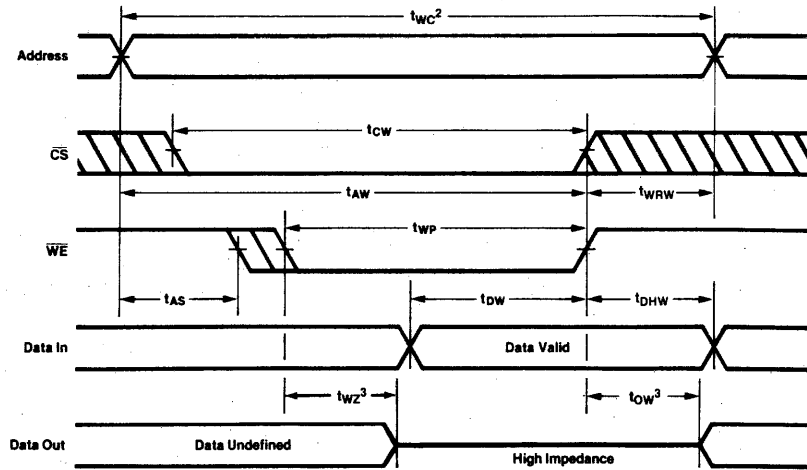


**TIMING WAVEFORM OF READ CYCLE NO. 2** <sup>(1,3)</sup>

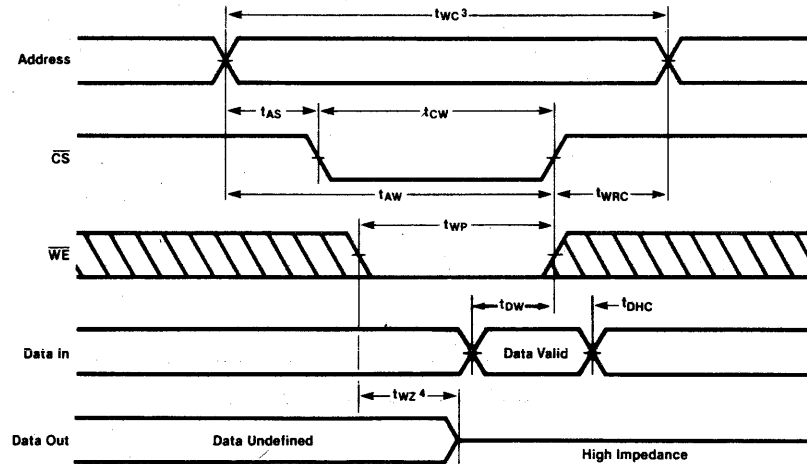


- NOTES:
- $\overline{WE}$  is high for READ cycle.
  - $\overline{CS}$  is low for READ cycle.
  - Address valid prior to or coincident with  $\overline{CS}$  transition low.
  - Transition is measured  $\pm 500mV$  from steady state voltage with specified loading in Figure 2. This parameter is sampled, not 100% tested.
  - All READ cycle timings are referenced from the last valid address to the first transitioning address.

**TIMING WAVEFORM OF WRITE CYCLE NO. 1 ( $\overline{WE}$  CONTROLLED)**



**TIMING WAVEFORM OF WRITE CYCLE NO. 2 ( $\overline{CS}$  CONTROLLED)**



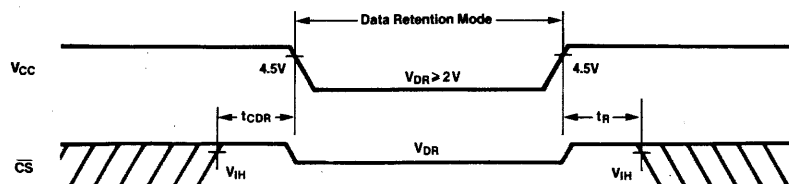
- NOTES: 1.  $\overline{WE}$  must be high during address transitions.  
 2. All write cycle timings are referenced from the last valid address to the first transitioning address.  
 3. Transition is measured  $\pm 100$ mV from steady state voltage with specified loading in Figure 2. This parameter is sampled and not 100% tested.

**LOW  $V_{CC}$  DATA RETENTION CHARACTERISTICS ( $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$  and  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ )**

SYMBOL	PARAMETER	TEST CONDITION	MIN.	TYP. <sup>(1)</sup>	MAX. COMM.	MAX. MIL.	UNIT
$V_{DR}$	$V_{CC}$ for Data Retention		2.0	—	—	—	V
$I_{CCDR}$	Data Retention Current	$\overline{CS} \geq V_{CC} - 0.2\text{V}$ $V_{IN} \geq V_{CC} - 0.2\text{V}$ or $\leq 0.2\text{V}$	—	2.0 <sup>(2)</sup>	60 <sup>(2)</sup>	800 <sup>(2)</sup>	$\mu\text{A}$
			—	4.0 <sup>(3)</sup>	90 <sup>(3)</sup>	1200 <sup>(3)</sup>	
$t_{CDR}$	Chip Deselect to Data Retention Time		0	—	—	—	ns
$t_R$	Operation Recovery Time		$t_{RC}$ <sup>(4)</sup>	—	—	—	ns

- NOTES: 1.  $T_A = 25^\circ\text{C}$   
 2. at  $V_{CC} = 2\text{V}$   
 3. at  $V_{CC} = 3\text{V}$   
 4.  $t_{RC}$  = Read Cycle Time

**LOW  $V_{CC}$  DATA RETENTION WAVEFORM**





Integrated Device Technology, Inc.

# 16x16 BIT PARALLEL CMOS MULTIPLIER

# IDT7216 IDT7217

## MILITARY / COMMERCIAL TEMPERATURE RANGES

### FEATURES:

- 16x16 parallel-multiplier with double precision product
- High-speed 75ns maximum clock to multiply time
- Produced with advanced CEMOS™ high-performance technology
- Low power consumption - 150mW typical, less than 1/10th the power of compatible bipolar parts
- IDT7216 is pin and functionally compatible with TRW MPY-16HJ and AMD AM29516
- IDT7217 requires only single clock with register enables making it pin and functionally compatible with AMD AM29517
- Configured for easy array expansion
- User controlled option for transparent output register mode
- Round control for rounding the MSP
- Single 5V power supply
- Input and output directly TTL-compatible
- Three-state output
- Available in DIP, LCC or Flat-Pak
- Standard product 100% screened to MIL-STD-883 Class C
- Military product available 100% screened to Class B

It has achieved speeds comparable to bipolar (75ns max.) at 1/10th the power consumption.

The IDT7216/IDT7217 are ideal for applications requiring high speed multiplications such as fast Fourier transform analysis, digital filtering, graphic display systems, speech synthesis and recognition, and in any system requirement where multiplication speeds of a mini/micro computer are inadequate.

All input registers, as well as LSP and MSP output registers, use the same positive edge triggered D type flip-flop. In the IDT7216, there are independent clocks (CLKX, CLKY, CLKM, CLKL) associated with each of these registers. The IDT7217 has only a single clock input (CLK) and three register enables. ENX and ENY control the two input registers, while ENP controls the entire product.

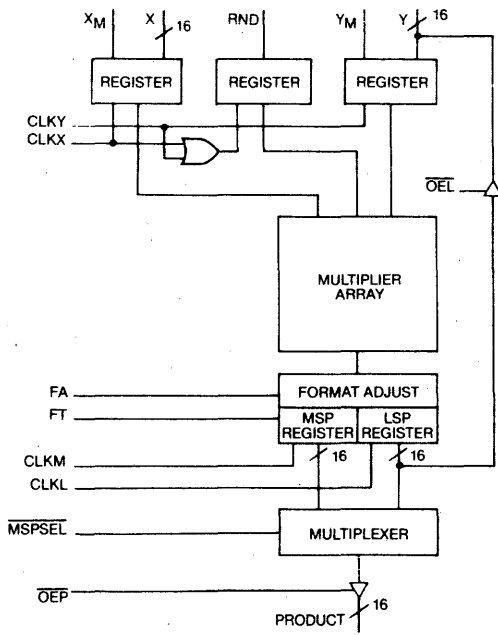
The IDT7216/7217 offer additional flexibility with the FA control and MSPSEL functions. The FA control formats the output for 2's complement by shifting the MSP up one bit and then repeating the sign bit in the MSP of the LSP. The MSPSEL low selects the MSP to be available at the product output port, while a high selects the LSP to be available. Keeping this pin low will ensure total compatibility with the TRW MPY-16HJ.

The IDT7216/IDT7217 Multipliers are 100% processed in compliance to the test methods of MIL-STD-883, Method 5004, making them ideally suited to applications demanding the highest level of performance and reliability.

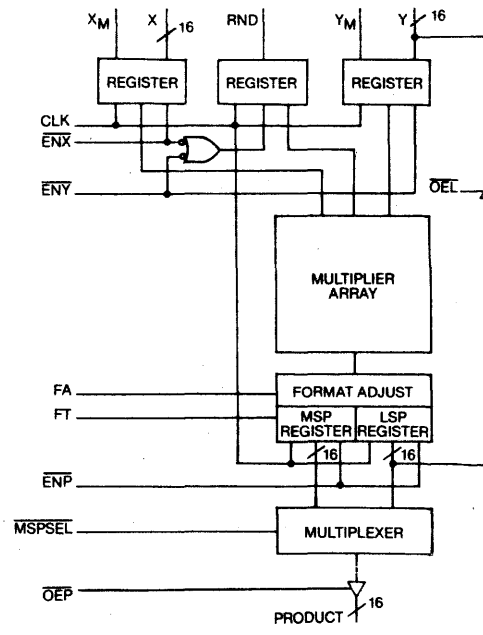
### DESCRIPTION:

The IDT7216/IDT7217 are high speed, low power 16x16 multipliers, ideal for fast, real time digital signal processing applications. Utilization of a modified Booths algorithm and IDT's high-performance, high reliability technology - CEMOS

### LOGIC DIAGRAMS



IDT7216



IDT7217

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Integrated Device Technology

MEMORY



ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

SYMBOL	RATING	COMMERCIAL	MILITARY	UNIT
V <sub>TERM</sub>	Terminal Voltage with Respect to GND	-0.5 to +7.0	-0.5 to +7.0	V
T <sub>A</sub>	Operating Temperature	0 to +70	-55 to +125	°C
T <sub>BIAS</sub>	Temperature Under Bias	-10 to +85	-65 to +135	°C
T <sub>STG</sub>	Storage Temperature	-55 to +125	-65 to +150	°C
P <sub>T</sub>	Power Dissipation	1.0	1.0	W
I <sub>OUT</sub>	DC Output Current	50	50	mA

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub> = 5V ± 5%, T<sub>A</sub> = 0°C to +70°C and V<sub>CC</sub> = 5V ± 10%, T<sub>A</sub> = -55°C to +125°C)

SYMBOL	PARAMETER	TEST CONDITIONS	T <sub>A</sub> = 0°C to 70°C			T <sub>A</sub> = -55°C to +125°C			UNIT
			MIN	TYP <sup>(1)</sup>	MAX	MIN	TYP <sup>(1)</sup>	MAX	
I <sub>LI</sub>	Input Leakage Current	V <sub>CC</sub> = Max, V <sub>IN</sub> = 0 to V <sub>CC</sub>	—	—	2	—	—	10	μA
I <sub>LO</sub>	Output Leakage Current	V <sub>OUT</sub> = 0 to V <sub>CC</sub>	—	—	2	—	—	10	μA
I <sub>CC</sub>	Operating Power Supply Current	Output Open	—	30	60	—	30	80	mA
I <sub>CCQ1</sub>	Quiescent Power Supply Current	V <sub>IN</sub> ≥ V <sub>IH</sub> or ≤ V <sub>IL</sub>	—	10	30	—	10	30	mA
I <sub>CCQ2</sub>	Quiescent Power Supply Current	V <sub>IN</sub> ≥ V <sub>CC</sub> - .2V or ≤ .2V	—	0.1	1.0	—	0.1	1.0	mA
V <sub>OH</sub>	Output High Voltage	V <sub>CC</sub> = Min., I <sub>OH</sub> = -0.4mA	2.4	—	—	2.4	—	—	V
V <sub>OL</sub>	Output Low Voltage	V <sub>CC</sub> = Min., I <sub>OL</sub> = 4.0mA	—	—	0.5	—	—	0.5	V

<sup>(1)</sup>V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.

## RECOMMENDED DC OPERATING CONDITIONS

Symbol	Parameter	MIN	TYP	MAX	UNIT
V <sub>CCM</sub>	Military Supply Voltage	4.5	5.0	5.5	V
V <sub>CCC</sub>	Commercial Supply Voltage	4.75	5.0	5.25	V
GND	Supply Voltage	0	0	0	V
V <sub>IH</sub>	Input High Voltage	2.0	—	—	V
V <sub>IL</sub>	Input Low Voltage	—	—	0.8	V

## TEST LOADS

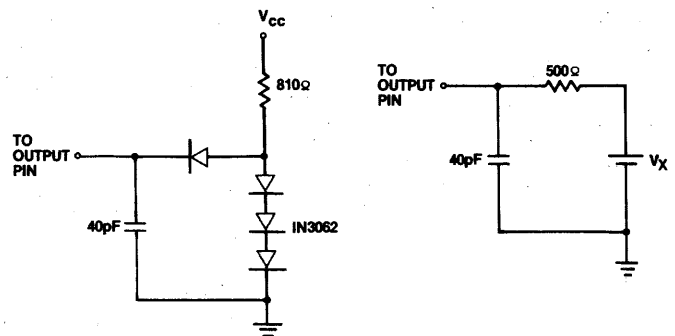


Figure 1. A.C. Output Test Load

Figure 2. Output Three State Delay Load

V<sub>X</sub> = 0V or 2.6V

## AC TEST CONDITIONS

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	5ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	See Figures 1 and 2

CAPACITANCE (T<sub>A</sub> = 25°C, f = 1.0MHz)

SYMBOL	ITEM	CONDITIONS	MAX.	UNIT
C <sub>IN</sub>	Input Capacitance	V <sub>IN</sub> = 0V	10	pF
C <sub>OUT</sub>	Output Capacitance	V <sub>OUT</sub> = 0V	12	pF

NOTE: This parameter is sampled and not 100% tested.

**AC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 5\%$ ,  $T_A = 0^\circ C$  to  $+70^\circ C$ ) **COMMERCIAL**

SYMBOL	PARAMETER	IDT7216-75 IDT7217-75		IDT7216-90 IDT7217-90		IDT7216-140 IDT7217-140		UNIT	TEST LOAD FIG.
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.		
$t_{MUC}$	Unlocked Multiply Time	—	110	—	125	—	180	ns	1
$t_{MC}$	Clocked Multiply Time	—	75	—	90	—	140	ns	1
$t_S$	X, Y, RND Setup Time	25	—	25	—	25	—	ns	1
$t_H$	X, Y, RND Hold Time	0	—	0	—	0	—	ns	1
$t_{PWH}$	Clock Pulse Width High	20	—	20	—	25	—	ns	1
$t_{PWL}$	Clock Pulse Width Low	20	—	20	—	25	—	ns	1
$t_{PDSEL}$	MSPSEL to Product Out	—	35	—	35	—	40	ns	1
$t_{PDP}$	Output Clock to P	—	35	—	35	—	40	ns	1
$t_{PDY}$	Output Clock to Y	—	35	—	35	—	40	ns	1
$t_{ENA}$	3 state enable time (Note 2)	—	35	—	35	—	40	ns	2
$t_{DIS}$	3 state disable time (Note 2)	—	30	—	30	—	40	ns	2
$t_S$	Clock Enable Setup Time (IDT7217 only)	25	—	25	—	25	—	ns	1
$t_H$	Clock Enable Hold Time (IDT7217 only)	0	—	0	—	0	—	ns	1
$t_{HCL}$	Clock Low Hold Time CLKXY Relative to CLKML (see Note 1) (IDT7216 only)	0	—	0	—	0	—	ns	1

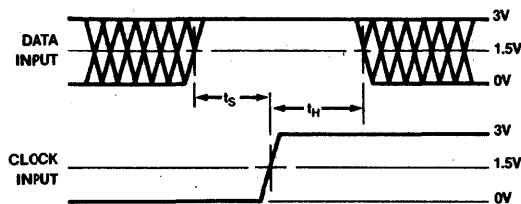
**AC ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5V \pm 10\%$ ,  $T_A = -55^\circ C$  to  $+125^\circ C$ ) **MILITARY**

SYMBOL	PARAMETER	IDT7216-90 IDT7217-90		IDT7216-120 IDT7217-120		IDT7216-185 IDT7217-185		UNIT	TEST LOAD FIG.
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.		
$t_{MUC}$	Unlocked Multiply Time	—	130	—	160	—	230	ns	1
$t_{MCS}$	Clocked Multiply Time	—	90	—	120	—	185	ns	1
$t_S$	X, Y, RND Setup Time	30	—	30	—	30	—	ns	1
$t_H$	X, Y, RND Hold Time	0	—	0	—	0	—	ns	1
$t_{PWH}$	Clock Pulse Width High	30	—	30	—	30	—	ns	1
$t_{PWL}$	Clock Pulse Width Low	30	—	30	—	30	—	ns	1
$t_{PDSEL}$	MSPSEL to Product Out	—	40	—	40	—	45	ns	1
$t_{PDP}$	Output Clock to P	—	40	—	40	—	45	ns	1
$t_{PDY}$	Output Clock to Y	—	40	—	40	—	45	ns	1
$t_{ENA}$	3 state enable time (Note 2)	—	40	—	40	—	45	ns	2
$t_{DIS}$	3 state disable time (Note 2)	—	40	—	40	—	45	ns	2
$t_S$	Clock Enable Setup Time (IDT7217 only)	30	—	30	—	30	—	ns	1
$t_H$	Clock Enable Hold Time (IDT7217 only)	0	—	0	—	0	—	ns	1
$t_{HCL}$	Clock Low Hold Time CLKXY Relative to CLKML (see Note 1) (IDT7216 only)	0	—	0	—	0	—	ns	1

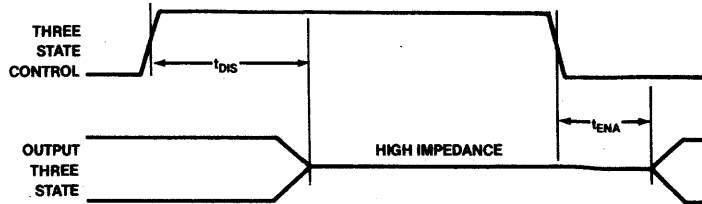
Note 1. To ensure that the correct product is entered in the output registers, new data may not be entered into the input registers before the output registers have been clocked.

2. Transition is measured  $\pm 500mV$  from steady state voltage with loading specified in Fig. 2

**SET-UP AND HOLD TIME**



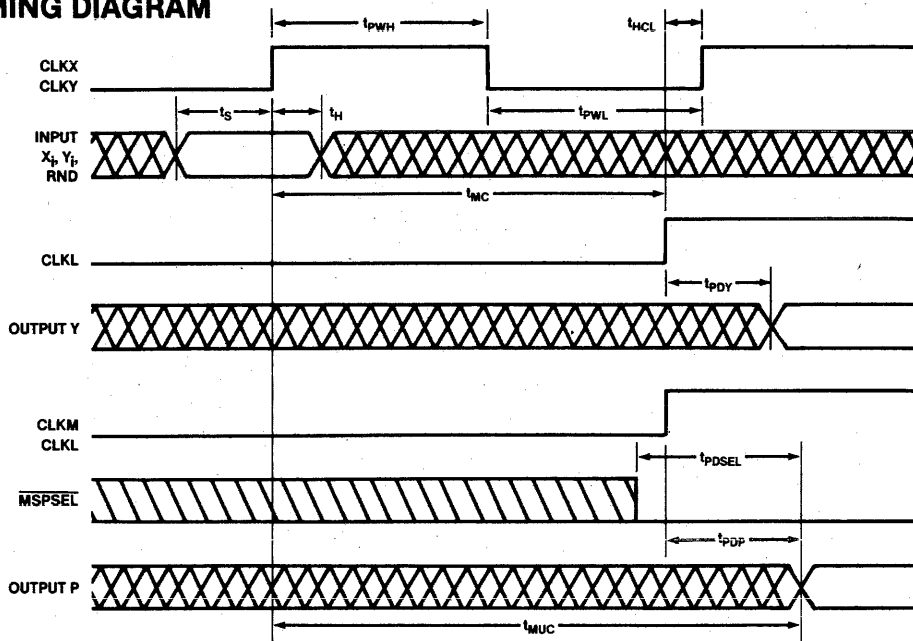
**THREE STATE CONTROL TIMING DIAGRAM**



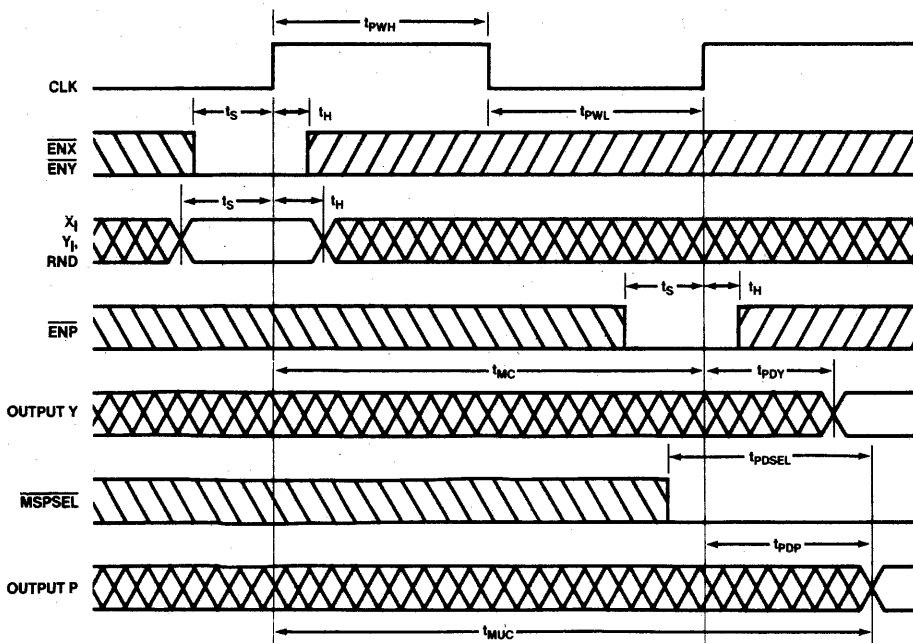
NOTES: Diagram shown for HIGH data only. Output transition may be opposite sense.

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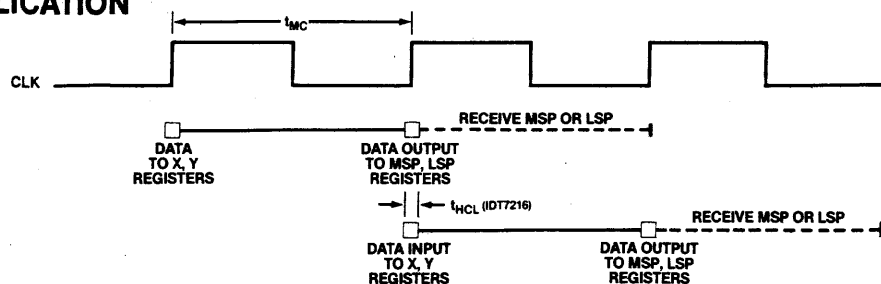
**IDT7216 TIMING DIAGRAM**



**IDT7217 TIMING DIAGRAM**



**SIMPLIFIED TIMING DIAGRAM  
TYPICAL APPLICATION**



**DEFINITION OF TERMS**

**16x16 BIT MULTIPLIER CONNECTION DIAGRAMS**

$X_{15}-X_0$	Multiplicand Data inputs.
$Y_{15}-Y_0$	Multiplier Data inputs.
$X_M, Y_M$ (TCX, TCY)*	Mode control inputs for each data word; LOW for unsigned data and HIGH for two's complement data.
FA (RS)*	Format adjust control selects either a full 32-bit product (HIGH) or a left shifted 31-bit product with the sign bit replicated in the LSP (LOW). This control is normally high, except for certain fractional two's complement applications. (See Multiplier output formats table).
FT	Feedthrough control (HIGH) makes both MSP and LSP registers transparent.
<u>MSPSEL</u>	Selects either MSP (LOW) or LSP (HIGH) to be available at the product output port.
RND	Round control for the rounding of the MSP.
<u>OE</u> P (TRIM)*	Three-state enable for product output port.
<u>OE</u> L (TRIL)*	Three-state enable for routing LSP through Y input/output port.
IDT7216 ONLY	
CLKX	Register Clock, $X_{15}-X_0, X_M, RND$
CLKY	Register Clock, $Y_{15}-Y_0, Y_M, RND$
CLKM	MSP Register Clock
CLKL	LSP Register Clock
IDT7217 ONLY	
CLK	Clock, All Registers
<u>EN</u> X	Register Enable, $X_{15}-X_0, X_M, RND$
<u>EN</u> Y	Register Enable, $Y_{15}-Y_0, Y_M, RND$
<u>EN</u> P	Register Enable, MSP, LSP

\* TRW MPY 16HJ pin designation.

PIN#	7216 DIP	7217 DIP	7216 FP	7217 FP	7216 LCC	7217 LCC
1	$X_4$	$X_4$	$P_{15} P_{31}$	$P_{15} P_{31}$	+V <sub>CC</sub>	+V <sub>CC</sub>
2	$X_3$	$X_3$	$P_{14} P_{30}$	$P_{14} P_{30}$	GND	GND
3	$X_2$	$X_2$	$P_{13} P_{29}$	$P_{13} P_{29}$	GND	GND
4	$X_1$	$X_1$	$P_{12} P_{28}$	$P_{12} P_{28}$	<u>MSPSEL</u>	<u>MSPSEL</u>
5	$X_0$	$X_0$	$P_{11} P_{27}$	$P_{11} P_{27}$	FT	FT
6	<u>OE</u> L	<u>OE</u> L	$P_{10} P_{26}$	$P_{10} P_{26}$	FA	FA
7	CLKL	CLK	$P_9 P_{25}$	$P_9 P_{25}$	<u>OE</u> P	<u>OE</u> P
8	CLKY	<u>EN</u> Y	$P_8 P_{24}$	$P_8 P_{24}$	CLKM	<u>EN</u> P
9	$P_0 Y_0$	$P_0 Y_0$	$P_7 P_{23}$	$P_7 P_{23}$	N/C	N/C
10	$P_1 Y_1$	$P_1 Y_1$	$P_6 P_{22}$	$P_6 P_{22}$	$P_{15} P_{31}$	$P_{15} P_{31}$
11	$P_2 Y_2$	$P_2 Y_2$	$P_5 P_{21}$	$P_5 P_{21}$	$P_{14} P_{30}$	$P_{14} P_{30}$
12	$P_3 Y_3$	$P_3 Y_3$	$P_4 P_{20}$	$P_4 P_{20}$	$P_{13} P_{29}$	$P_{13} P_{29}$
13	$P_4 Y_4$	$P_4 Y_4$	$P_3 P_{19}$	$P_3 P_{19}$	$P_{12} P_{28}$	$P_{12} P_{28}$
14	$P_5 Y_5$	$P_5 Y_5$	$P_2 P_{18}$	$P_2 P_{18}$	$P_{11} P_{27}$	$P_{11} P_{27}$
15	$P_6 Y_6$	$P_6 Y_6$	$P_1 P_{17}$	$P_1 P_{17}$	$P_{10} P_{26}$	$P_{10} P_{26}$
16	$P_7 Y_7$	$P_7 Y_7$	$P_0 P_{16}$	$P_0 P_{16}$	$P_9 P_{25}$	$P_9 P_{25}$
17	$P_8 Y_8$	$P_8 Y_8$	$P_{15} Y_{15}$	$P_{15} Y_{15}$	$P_8 P_{24}$	$P_8 P_{24}$
18	$P_9 Y_9$	$P_9 Y_9$	$P_{14} Y_{14}$	$P_{14} Y_{14}$	$P_7 P_{23}$	$P_7 P_{23}$
19	$P_{10} Y_{10}$	$P_{10} Y_{10}$	$P_{13} Y_{13}$	$P_{13} Y_{13}$	$P_6 P_{22}$	$P_6 P_{22}$
20	$P_{11} Y_{11}$	$P_{11} Y_{11}$	$P_{12} Y_{12}$	$P_{12} Y_{12}$	$P_5 P_{21}$	$P_5 P_{21}$
21	$P_{12} Y_{12}$	$P_{12} Y_{12}$	$P_{11} Y_{11}$	$P_{11} Y_{11}$	$P_4 P_{20}$	$P_4 P_{20}$
22	$P_{13} Y_{13}$	$P_{13} Y_{13}$	$P_{10} Y_{10}$	$P_{10} Y_{10}$	$P_3 P_{19}$	$P_3 P_{19}$
23	$P_{14} Y_{14}$	$P_{14} Y_{14}$	$P_9 Y_9$	$P_9 Y_9$	$P_2 P_{18}$	$P_2 P_{18}$
24	$P_{15} Y_{15}$	$P_{15} Y_{15}$	$P_8 Y_8$	$P_8 Y_8$	$P_1 P_{17}$	$P_1 P_{17}$
25	$P_0 P_{16}$	$P_0 P_{16}$	$P_7 Y_7$	$P_7 Y_7$	$P_0 P_{16}$	$P_0 P_{16}$
26	$P_1 P_{17}$	$P_1 P_{17}$	$P_6 Y_6$	$P_6 Y_6$	N/C	N/C
27	$P_2 P_{18}$	$P_2 P_{18}$	$P_5 Y_5$	$P_5 Y_5$	$P_{15} Y_{15}$	$P_{15} Y_{15}$
28	$P_3 P_{19}$	$P_3 P_{19}$	$P_4 Y_4$	$P_4 Y_4$	$P_{14} Y_{14}$	$P_{14} Y_{14}$
29	$P_4 P_{20}$	$P_4 P_{20}$	$P_3 Y_3$	$P_3 Y_3$	$P_{13} Y_{13}$	$P_{13} Y_{13}$
30	$P_5 P_{21}$	$P_5 P_{21}$	$P_2 Y_2$	$P_2 Y_2$	$P_{12} Y_{12}$	$P_{12} Y_{12}$
31	$P_6 P_{22}$	$P_6 P_{22}$	$P_1 Y_1$	$P_1 Y_1$	$P_{11} Y_{11}$	$P_{11} Y_{11}$
32	$P_7 P_{23}$	$P_7 P_{23}$	$P_0 Y_0$	$P_0 Y_0$	$P_{10} Y_{10}$	$P_{10} Y_{10}$
33	$P_8 P_{24}$	$P_8 P_{24}$	CLKY	<u>EN</u> Y	$P_9 Y_9$	$P_9 Y_9$
34	$P_9 P_{25}$	$P_9 P_{25}$	CLKL	CLK	$P_8 Y_8$	$P_8 Y_8$
35	$P_{10} P_{26}$	$P_{10} P_{26}$	<u>OE</u> L	<u>OE</u> L	$P_7 Y_7$	$P_7 Y_7$
36	$P_{11} P_{27}$	$P_{11} P_{27}$	$X_0$	$X_0$	$P_6 Y_6$	$P_6 Y_6$
37	$P_{12} P_{28}$	$P_{12} P_{28}$	$X_1$	$X_1$	$P_5 Y_5$	$P_5 Y_5$
38	$P_{13} P_{29}$	$P_{13} P_{29}$	$X_2$	$X_2$	$P_4 Y_4$	$P_4 Y_4$
39	$P_{14} P_{30}$	$P_{14} P_{30}$	$X_3$	$X_3$	$P_3 Y_3$	$P_3 Y_3$
40	$P_{15} P_{31}$	$P_{15} P_{31}$	$X_4$	$X_4$	$P_2 Y_2$	$P_2 Y_2$
41	CLKM	<u>EN</u> P	$X_5$	$X_5$	$P_1 Y_1$	$P_1 Y_1$
42	<u>OE</u> P	<u>OE</u> P	$X_6$	$X_6$	$P_0 Y_0$	$P_0 Y_0$
43	FA	FA	$X_7$	$X_7$	N/C	N/C
44	FT	FT	$X_8$	$X_8$	CLKY	<u>EN</u> Y
45	<u>MSPSEL</u>	<u>MSPSEL</u>	$X_9$	$X_9$	CLKL	CLK
46	GND	GND	$X_{10}$	$X_{10}$	<u>OE</u> L	<u>OE</u> L
47	GND	GND	$X_{11}$	$X_{11}$	$X_0$	$X_0$
48	+V <sub>CC</sub>	+V <sub>CC</sub>	$X_{12}$	$X_{12}$	$X_1$	$X_1$
49	+V <sub>CC</sub>	+V <sub>CC</sub>	$X_{13}$	$X_{13}$	$X_2$	$X_2$
50	$Y_M$	$Y_M$	$X_{14}$	$X_{14}$	$X_3$	$X_3$
51	$X_M$	$X_M$	$X_{15}$	$X_{15}$	$X_4$	$X_4$
52	RND	RND	CLKX	<u>EN</u> X	$X_5$	$X_5$
53	CLKX	<u>EN</u> X	RND	RND	$X_6$	$X_6$
54	$X_{15}$	$X_{15}$	$X_M$	$X_M$	$X_7$	$X_7$
55	$X_{14}$	$X_{14}$	$Y_M$	$Y_M$	$X_8$	$X_8$
56	$X_{13}$	$X_{13}$	+V <sub>CC</sub>	+V <sub>CC</sub>	$X_9$	$X_9$
57	$X_{12}$	$X_{12}$	+V <sub>CC</sub>	+V <sub>CC</sub>	$X_{10}$	$X_{10}$
58	$X_{11}$	$X_{11}$	GND	GND	$X_{11}$	$X_{11}$
59	$X_{10}$	$X_{10}$	GND	GND	$X_{12}$	$X_{12}$
60	$X_9$	$X_9$	<u>MSPSEL</u>	<u>MSPSEL</u>	N/C	N/C
61	$X_8$	$X_8$	FT	FT	$X_{13}$	$X_{13}$
62	$X_7$	$X_7$	FA	FA	$X_{14}$	$X_{14}$
63	$X_6$	$X_6$	<u>OE</u> P	<u>OE</u> P	$X_{15}$	$X_{15}$
64	$X_5$	$X_5$	CLKM	<u>EN</u> P	CLKX	<u>EN</u> X
65	—	—	—	—	RND	RND
66	—	—	—	—	$X_M$	$X_M$
67	—	—	—	—	$Y_M$	$Y_M$
68	—	—	—	—	+V <sub>CC</sub>	+V <sub>CC</sub>

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**IDT MILITARY PROCESSING TO MIL-STD-883**

Maintaining the highest standards of quality in our memory products is the basis of IDT's standard manufacturing systems and procedures. IDT products begin with stringent design rules derived for use in high reliability programs. This is followed by a dedicated commitment to reliable workmanship as well as rigid controls throughout wafer fab, device assembly and electrical test, all of which are designed to produce products that are inherently reliable.

All monolithic memory products, including commercial grade, are manufactured and screened to the demanding requirements of MIL-STD-883, Method 5004, Class C. Documentation, design, processing and assembly workmanship guidelines for these monolithic components are patterned after MIL-M-38510 specifications.

For customer applications requiring higher levels of reliability screening, we supply full military range monolithic components completely screened to the Class B

criteria of Method 5004. This includes 100% 160-hour burn-in at  $T_A = +125^\circ\text{C}$  (or equivalent) per Method 1015 followed by 100% temperature testing of all DC and AC parameters and DC functional characteristics over the full  $-55^\circ\text{C}$  to  $+125^\circ\text{C}$  temperature range.

In the case of module assemblies, additional screening of the fully assembled RAMPACKs is performed per Figure 3 to assure package integrity and mechanical reliability. Finally, 100% electrical tests are performed.

Samples of the monolithic product which have been processed to Method 5004 100% screening requirements are submitted to the Quality Conformance inspection requirements of MIL-STD-883. These Quality Conformance inspections, as shown in Figure 2, are performed to the criteria of Method 5005, Group A (electrical), Group B (mechanical), Group C (chip integrity), and Group D (package environmental integrity).

For special customer specifications or quality requirements beyond Class B levels of MIL-STD-883 — such as SEM analysis, X-ray, or other screening flows to meet specific needs — contact your local IDT sales office.

**MONOLITHIC COMPONENT SCREENING PROCEDURES**

PER MIL-STD-883, METHOD 5004, CLASS B

SCREEN	TEST METHOD	LEVEL
Visual and Mechanical Internal Visual	2010 Condition B	100%
High-Temperature Storage	1008, Condition C	100%
Temperature Cycle	1010, Condition C	100%
Constant Acceleration	2001	100%
Hermeticity, Fine and Gross	1014	100%
Burn-In Pre-Burn-In Electrical	Per Applicable Device Specification	100%
Burn-In	1015, 160 Hrs. @ $+125^\circ\text{C}$ or Equivalent	100%
Final Electrical Tests Static (DC)	a. At $25^\circ\text{C}$ and Power Supply Extremes	100%
	b. At Temperature and Power Supply Extremes	100%
Functional	a. At $25^\circ\text{C}$ and Power Supply Extremes	100%
	b. At Temperature and Power Supply Extremes (IDT imposed)	100%
Switching (AC) or Dynamic	a. At $25^\circ\text{C}$ and Power Supply Extremes	100%
	b. At Temperature and Power Supply Extremes (IDT imposed)	100%
External Visual	2009	100%

Figure 1

**QUALITY CONFORMANCE TESTING PER MIL-STD-883, METHOD 5005, CLASS B**

SCREEN	TEST METHOD	LEVEL
Quality Conformance Sample Tests	Group A (Electrical Tests)	Sample
	Group B (Mechanical Tests)	Sample
	Group C (Chip Integrity Tests)	Sample
	Group D (Package Integrity Tests)	Sample

Figure 2

**FULLY ASSEMBLED MODULE SCREENING PER MIL-STD-883, METHOD 5004, CLASS B**

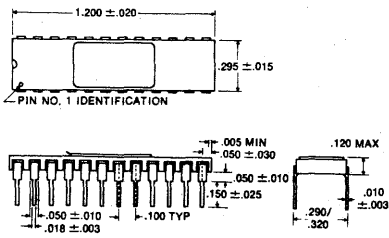
SCREEN	TEST METHOD	LEVEL
Temperature Cycle	1010 Condition C	100%
Hermeticity Fine and Gross	1014	100%
Final Electrical Tests Static (DC)	a. At $25^\circ\text{C}$ and Power Supply Extremes	100%
	b. At Temperature and Power Supply Extremes	100%
Functional	a. At $25^\circ\text{C}$ and Power Supply Extremes	100%
	b. At Temperature and Power Supply Extremes (IDT exposed)	100%
Switching (AC) or Dynamic	a. At $25^\circ\text{C}$ and Power Supply Extremes	100%
	b. At Temperature and Power Supply Extremes (IDT exposed)	100%
External Visual	2009	100%

Figure 3

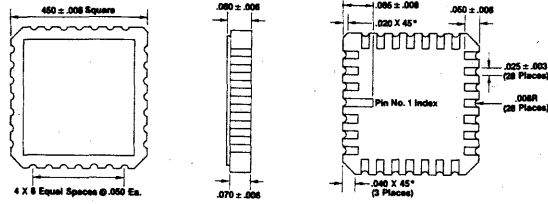
For special customer specifications or quality requirements beyond Class B levels of MIL-STD-883, such as SEM analysis, X-ray, or other screening flows to meet specific needs, contact your local IDT sales office.

CMOS PRODUCTS PACKAGE DIAGRAMS  
IDT6116/IDT71681/IDT71682

24-LEAD  
THINDIP  
SIDEBRAZE

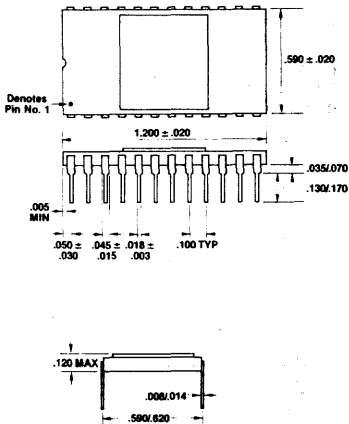


28 PIN LCC

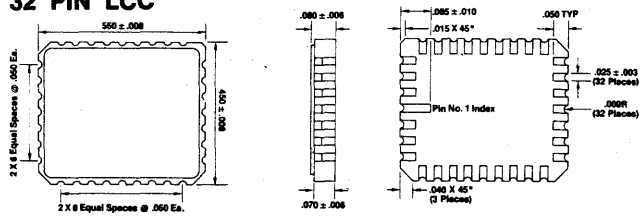


IDT6116

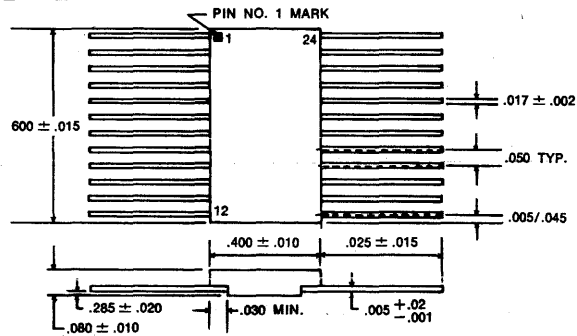
24-LEAD SIDEBRAZE



32 PIN LCC

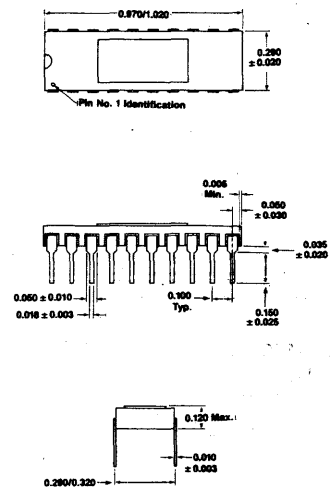


24-LEAD FLATPACK

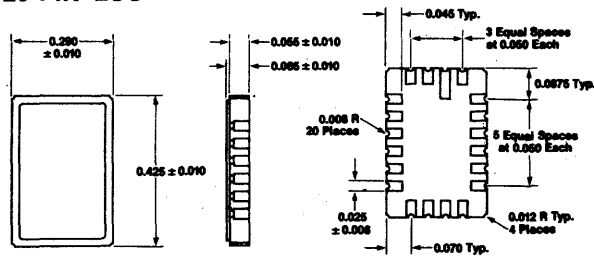


IDT6167/IDT6168

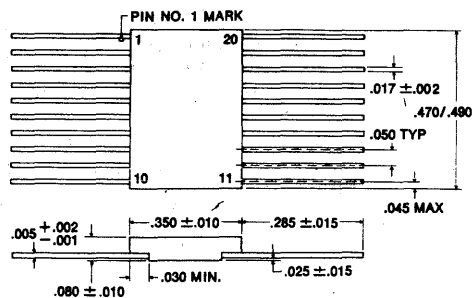
20-LEAD SIDEBRAZE



20-PIN LCC



20-LEAD FLATPACK

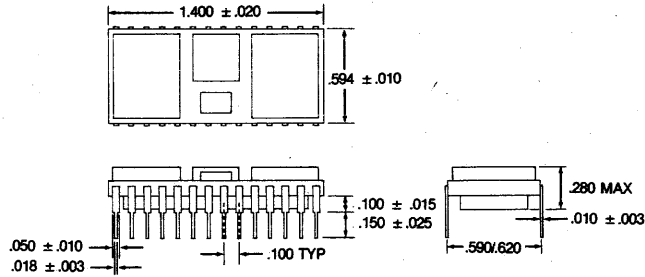


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CMOS PRODUCTS PACKAGE DIAGRAMS

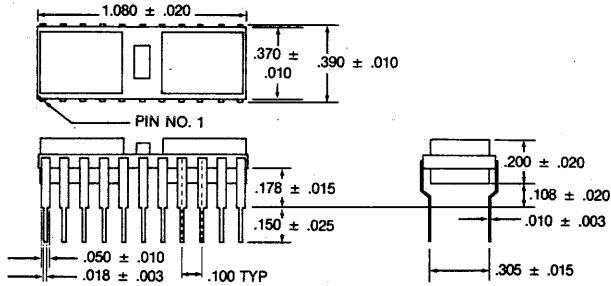
IDT7M/8M864

28-LEAD  
SIDEBRAZE



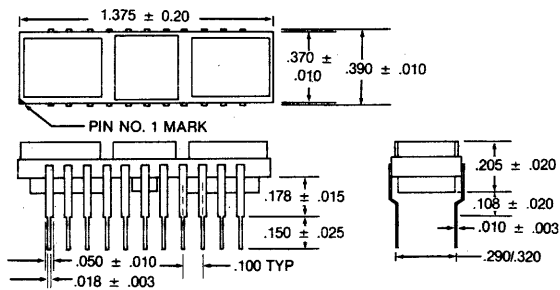
IDT7M464 (BENT LEADS)

22-LEAD  
SIDEBRAZE



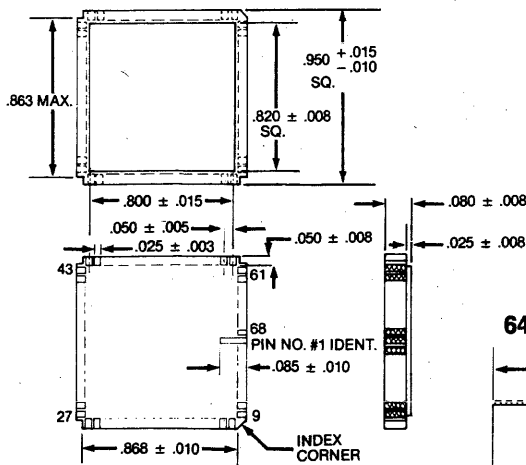
IDT7M164

22-LEAD  
SIDEBRAZE

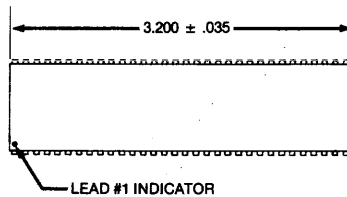


IDT7216/7217

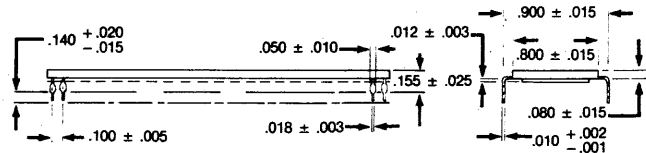
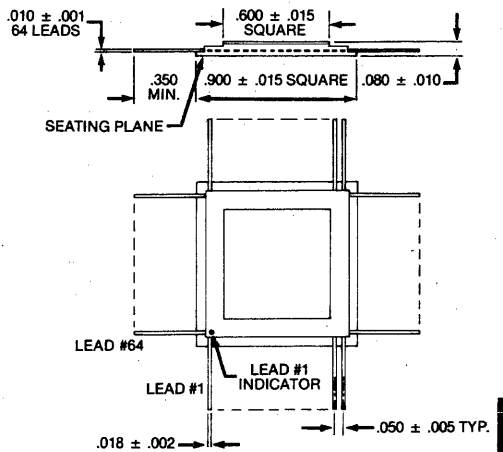
68-PIN LCC



64-LEAD TOPBRAZE DIP



64-LEAD FLATPACK



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## BPK 72A 1MBIT BUBBLE MEMORY PROTOTYPE KIT

Device	Case Operating Temperature
BPK 72A-1	0 To 75°C
BPK 72A-4	10°C To 55°C
BPK 72A-5	- 20°C To 85°C

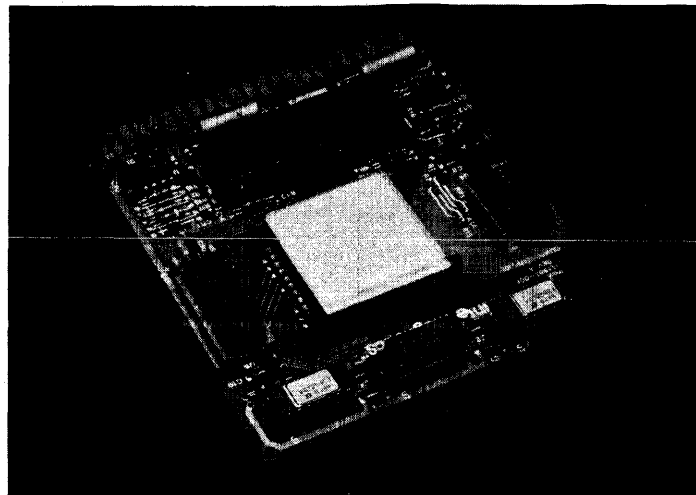
- Assembled and Tested 1MBit Bubble Memory Prototype Kit on 4" x 4" PC Board
- Complete with Powerfail Data Protection and Clock Circuitry
- Built-in Error Detection/Correction
- Interfaces with Intel 8080/85/86/88 186/286 and Other Microprocessors
- Software Driver for Bubble Memory Kit Diskette for Intel MDS\* System
- 1Mbit, Non-Volatile, Read-Write, Solid-State Memory in Leaded Dense Package
- Average Random Access Time of 48ms
- Maximum Data Rate of 100K bit/sec
- Operates from +5V and +12V Power Supplies
- Complete Documentation and Interfacing Information Included

The BPK 72A prototype kit is a completely assembled and tested 1Mbit bubble memory evaluation tool. It is ideal for the design engineer that wants the opportunity to fast evaluate how a bubble memory solution improves and adds value to an end-product by providing a compact solid-state memory that also reliably keeps the data at any powerdown.

Application information on microprocessor interfacing is included in the kit. A Bubble Memory Kit software driver is also included on a diskette for the Intel MDS\* System.

For production purposes, the bubble memory and the support circuit components can be ordered as the BPK 70A Subsystem. The 7220, controlling up to eight BPK 70A's, is ordered separately.

\*MDS is a registered trademark of Mohawk Data Science Corp.



## BPK 70A 1MBIT BUBBLE MEMORY SUBSYSTEMS

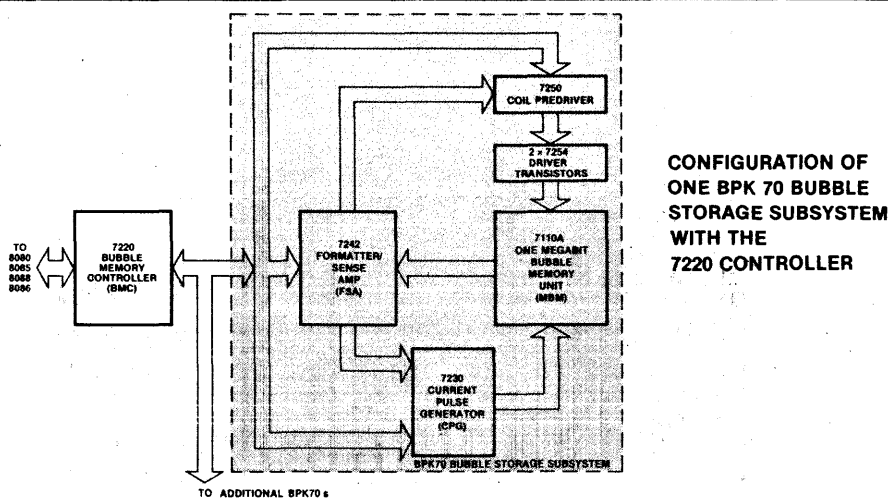
Device	Case Operating Temperature
BPK 70A-1	0 To 75°C
BPK 70A-4	10°C To 55°C
BPK 70A-5	- 20°C To 85°C

- 1 Mbit (128K Bytes) Non-Volatile, Solid-State, Read/Write Bubble Memory Subsystem
- Interfaces to Host Microprocessor Via Additional Bubble Memory Controller
- Maximum Data Rate of 100K Bit/Sec with One Subsystem
- Maximum Data Rate of 800K Bit/Sec with Eight Subsystems in Parallel
- Contains Bubble Memory and ICs for Production with 1Mbit Bubble Memory
- Modularity Provides Expansion Up to Eight Subsystems Per Controller
- Average Random Access Time of 48ms
- Bubble Memory in 20-Pin, Dual-In-Line Leaded Package

A BPK 70A Bubble Storage Subsystem is a modular building block consisting of one 1Mbit Magnetic Bubble Memory and five support ICs (Shown in the figure below). The MBM is in a leaded package requiring no socket. These are the basic components to build a non-volatile, solid-state, read-write memory system utilizing 1Mbit bubble memory.

An additional 1Mbit Bubble Memory Controller (BMC) 7220 provides the interface between the host microprocessor and the Bubble Memory Subsystem and provides all the timing and control signals to the Subsystem. The user interface of the BMC is compatible with microprocessor bus systems for 8080, 8085, 8086, 8088, 80186, 80286 and other standard microprocessors. The BMC is a software driven device utilizing 16 convenient commands. The design engineer's primary responsibility is the interfacing to the BMC. This interfacing is comparable to the one of a disk drive controller.

The modular Intel Subsystem provides a path for density expansion. One BMC can interface up to eight BPK 70A subsystems. Thus a 1Mbit (128K byte) system can be expanded up to 8Mbit (1M byte) by using additional subsystems.



CONFIGURATION OF ONE BPK 70 BUBBLE STORAGE SUBSYSTEM WITH THE 7220 CONTROLLER

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OCTOBER 1983  
ORDER NO. 210356-003

2004

4K (512 x 8) NON-VOLATILE RANDOM ACCESS MEMORY

- 5 Volt Only Operation
- Fast Static RAM Read/Write Cycles
  - 2004-2, 200ns Max.
  - 2004, 250ns Max.
  - 2004-3, 300ns Max.
- Single Line STORE & RECALL
- 10ms Self-Timed STORE Cycles for 2004-2 and 2004 (20ms for 2004-3)
- Automatic Recall on Power Up
- Write Protect Circuit to Preserve Data On Power-Up and Power-Down
- Lower Power Standby Mode
- 10-Year Data Retention for each STORE
- Minimum 10,000 Non-Volatile STORE Cycle Endurance
- Unlimited Endurance for Read, Write, and RECALL Cycles
- HMOS\*-E FLOTOX Cell Design
- Conforms to JEDEC Byte-Wide Universal Site

The Intel 2004 Non-Volatile Random Access Memory (NVRAM) is a 4K device with 512 x 8 architecture. It provides the real-time read/write functions of a static RAM together with the reliable non-volatile storage capability of an E<sup>2</sup>PROM array to preserve its memory contents when power is removed.

Internally, the 2004 NVRAM consists of a high speed static RAM array backed up, bit-for-bit, by an E<sup>2</sup>PROM array for non-volatile storage. The transfer of memory data between the static RAM and the E<sup>2</sup>PROM array occurs in parallel for fast storage and recall as well as minimal system support.

Two functions are provided to transfer data between the volatile RAM and its non-volatile E<sup>2</sup>PROM counterpart. The STORE function transfers RAM data into the E<sup>2</sup>PROM while the RECALL function fetches E<sup>2</sup>PROM data and places it in the RAM array. Both functions are controlled by a single  $\overline{NE}$  signal which can easily be activated with traditional circuitry in memory mapped space, through an I/O port, or from the output of a power-fail detector.

The RAM operating characteristics of the 2004 NVRAM provides high speed microprocessor performance with unlimited endurance. In the non-volatile storage mode, data retention is specified at over 10 years for each STORE operation. Over 10,000 STORE operations can be performed reliably.

The 2004 NVRAM is furnished in a 28-pin byte wide package with its address, data and control lines configured according to the standard JEDEC universal 28-pin site.

\*HMOS is patent process of Intel Corporation.

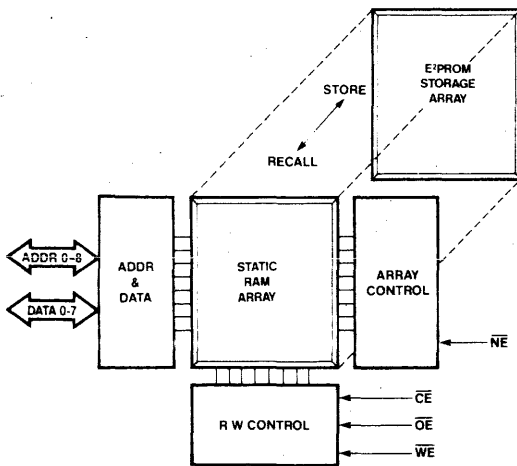


Figure 1. 2004 Functional Diagram

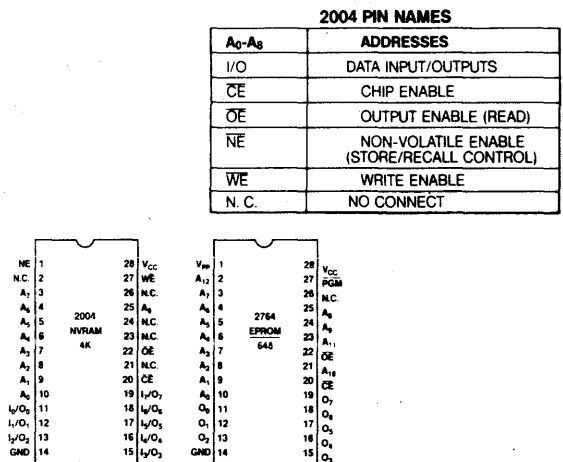


Figure 2. 2004 Pin Configuration

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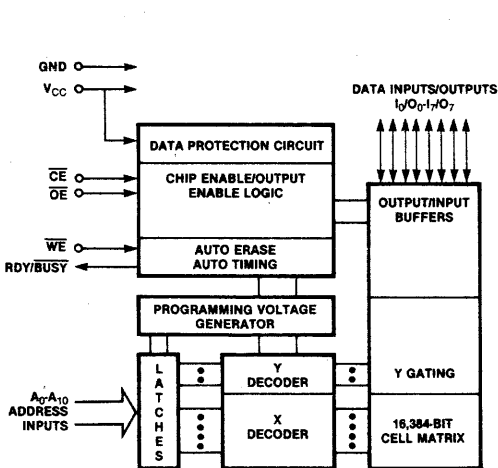
# 2817A

## 16K (2K x 8) ELECTRICALLY ERASABLE PROM

- 5 Volt Only Operation
- On-Chip Latches for Direct Microprocessor Interface
- Automatic Byte-Erase-before-Write
- Self Timed Byte Write
- Fast Read Access Time:
  - 2817A-1      200ns max
  - 2817A-2      200ns max
  - 2817A        250ns max
  - 2817A-3      350ns max
  - 2817A-4      450ns max
- Write Protect Circuit to Preserve Data on Power Up and Power Down
- 10,000 Erase/Write Cycles per Byte
- Reliable Intel HMOS\*-E FLOTOX Cell Design Technology
- $\overline{\text{RDY/BUSY}}$  Line for End-of-Write Signal
- 10 Year Data Retention For Each Write

The Intel 2817A is a 16,384 bit Electrically Erasable Programmable Read Only Memory. Like the Intel 2816A it has completely Non-Volatile Data Storage. In addition, it offers a high degree of integrated functionality which enables in-circuit byte writes to be performed with minimal hardware and software overhead. The Intel 2817A is a product of Intel's advanced E<sup>2</sup>PROM technology and uses the powerful HMOS\*-E process for reliable, non-volatile data storage.

\*HMOS is a patented process of Intel Corporation.



### PIN NAMES

A <sub>7</sub> -A <sub>10</sub>	ADDRESSES
CE	CHIP ENABLE
OE	OUTPUT ENABLE
O <sub>0</sub> -O <sub>7</sub>	DATA OUTPUTS
I <sub>0</sub> -I <sub>7</sub>	DATA INPUTS
RDY/BUSY	DEVICE READY/BUSY
N.C.	NO CONNECT
WE	WRITE ENABLE

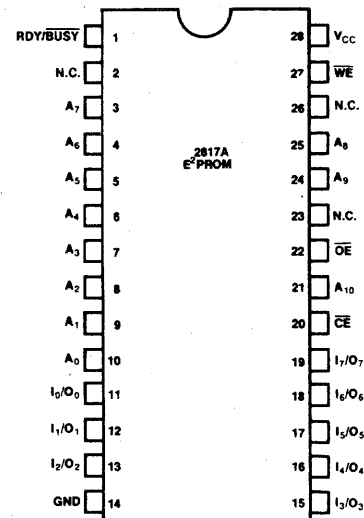


Figure 1. 2817A Functional Block Diagram

Figure 2. 2817A Pin Configuration

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OCTOBER 1983  
 ORDER NO. 230641-002

# 27256

## 256K (32K x 8) UV ERASABLE PROM

- Software Carrier Capability
- 250 ns Maximum Access Time
- Two-Line Control
- intelligent Identifier™ Mode
  - Automated Programming Operations
- TTL Compatible
- Industry Standard Pinout . . . JEDEC Approved
- Low Power
  - 100 mA max. Active
  - 40 mA max. Standby
- intelligent Programming™ Algorithm
  - Fastest EPROM Programming

The Intel 27256 is a 5V only, 262,144-bit ultraviolet Erasable and Electrically Programmable Read Only Memory (EPROM). Organized as 32K words by 8 bits, individual bytes are accessed in under 250ns. This is compatible with high performance microprocessors, such as the Intel 8MHz iAPX 186, allowing full speed operation without the addition of performance-degrading WAIT states. The 27256 is also directly compatible with Intel's 8051 family of microcontrollers.

The 27256 enables implementation of new, advanced systems with firmware intensive architectures. The combination of the 27256's high density, cost effective EPROM storage, and new advanced microprocessors having megabit addressing capability provides designers with opportunities to engineer user-friendly, high reliability, high-performance systems.

The 27256's large storage capability of 32K bytes enables it to function as a high density software carrier. Entire operating systems, diagnostics, high-level language programs and specialized application software can reside in a 27256 EPROM directly on a system's memory bus. This permits immediate microprocessor access and execution of software and eliminates the need for time consuming disk accesses and downloads.

Several advanced features have been designed into the 27256 that allow for fast and reliable programming—the intelligent identifier™ mode and the intelligent Programming™ Algorithm. Programming equipment that takes advantage of these innovations will electronically identify the 27256 and then rapidly program it using an efficient programming method.

Two-line control and JEDEC-approved, 28-pin packaging are standard features of all Intel high-density EPROMs. This assures easy microprocessor interfacing and minimum design efforts when upgrading, adding, or choosing between nonvolatile memory alternatives.

The 27256 is manufactured using Intel's advanced HMOS\*II-E technology.

\*HMOS is a patented process of Intel Corporation.

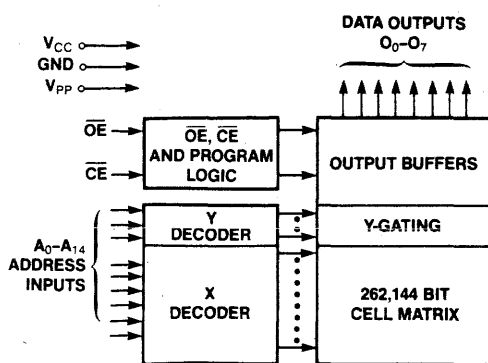
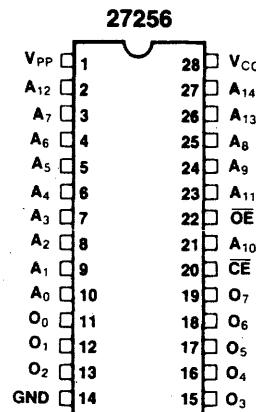


Figure 1. Block Diagram



PIN NAMES	
A <sub>0</sub> -A <sub>14</sub>	ADDRESSES
CE	CHIP ENABLE
OE	OUTPUT ENABLE
O <sub>0</sub> -O <sub>7</sub>	OUTPUTS

Figure 2. Pin Configuration

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JUNE 1983

ORDER NUMBER: 210827-004

# High Performance Ti-W PROM Family

## 53/63SXXX 53/63SXXXA

### Features/Benefit

- From 256 Bit to 32768 Bit of memory
- Reliable Titanium-Tungsten fuses (Ti-W)
- Low voltage programming
- Highest speed Schottky PROM family available
- Pin compatible with standard Schottky PROMs
- PNP inputs for low input current
- Compatible pin configurations for upward expansion

### Applications

- Microprogram control store
- microprocessor program store
- Look up table
- Character generator
- Random logic
- Code converter

### Description

The family features common electrical parameters and programming algorithm, low input current PNP inputs, full Schottky clamping and three-state and open collector outputs. The titanium-tungsten fuses store a logical low and are programmed to the high state. Special on chip circuitry and extra fuses provide preprogramming testing which assure high programming yields and high reliability.

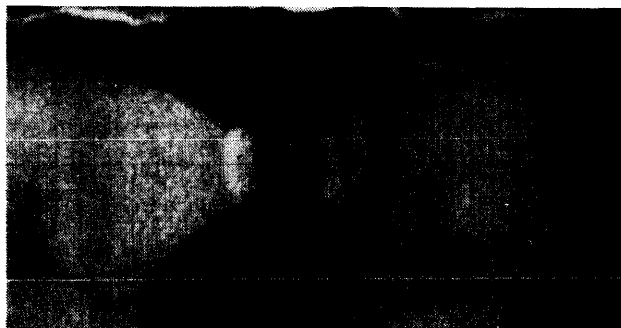
The 63 series is specified for operation over the commercial temperature and voltage range. the 53 series is specified for the military ranges.

### New Programming Technique:

Our new HIGH Performance PROMs use an elevated voltage at  $V_{CC}$  instead of using a separate programming pin (one of the enables) as in the Standard Performance PROMs using nichrome fuses. Changes in the internal circuitry were made to optimize speed and accordingly the unblown fuse represents a LOW at the output. When a fuse is programmed it reflects a high at the output.



Unblown Fuse



Blown Fuse

### High Performance PROM Selection Guide

MEMORY			PACKAGE		DEVICE TYPE	
SIZE	ORGANIZATION		PINS	TYPE	0° C to + 75° C	- 55° C to + 125° C
¼K	32 x 8	OC	16, (20)	N,J,F,W (L)	63S080	53S080
		TS			63S081	53 S081
1K	256 x 4	OC	16, (20)	N,J,F,W (L)	63S140	53S140
		TS			63S141	53S141
2K	512 x 4	OC	16, (20)	N,J,F,W (L)	63S240	53S240
		TS			63S241	53S241
4K	1024 x 4	TS	18, (20)	N,J,F, (L)	63S441	53S441
					63S441A	53S441A
8K	2048 x 4	TS	18, (28)	N,J,F, (L)	63S841	53S841
					63S841A	53S841A
16K	4096 x 4	TS	20	N, J, F	63S1641	53S1641
					63S1641A	53S1641A
	2048 x 8	TS	24, (28)	J,JS,F, (L)	63S1681	53S1681
					63S1681A	53S1681A
32K	4096 x 8	TS	24, (28)	*J (L)	63S3281	53S3281
					63S3281A	53S3281A

\* Flat-pack contact the factory ( ) = Military Product

MEMORY

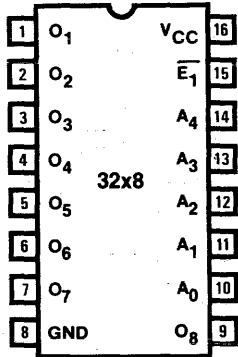
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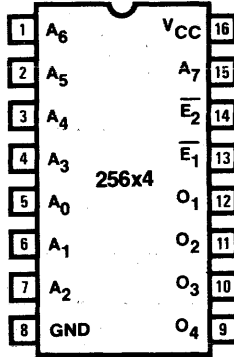
**Monolithic Memories** 

Pin Configurations

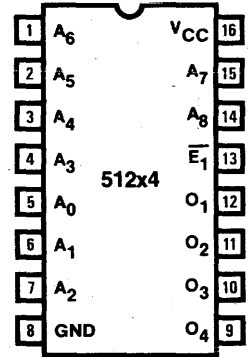
53/63S080  
53/63S081



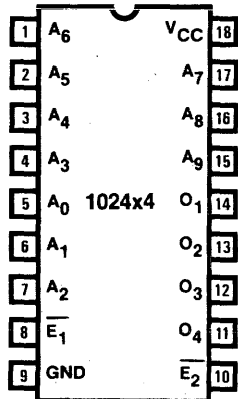
53/63S140  
53/63S141



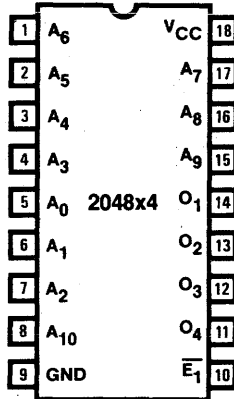
53/63S240  
53/63S241



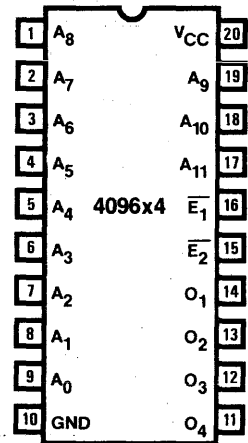
53/63S441  
53/63S441A



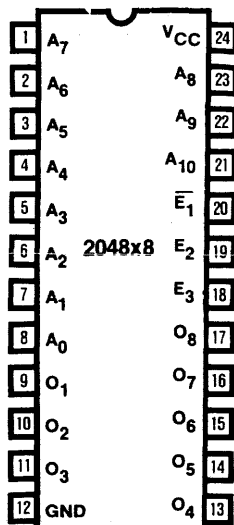
53/63S841  
53/63S841A



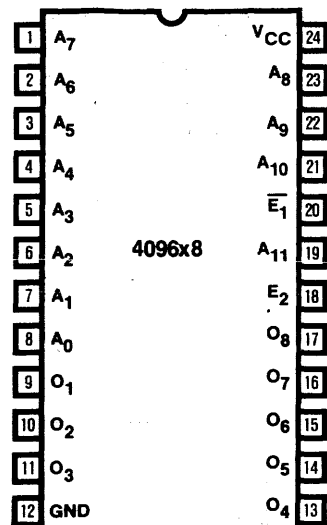
53/63S1641  
53/63S1641A



53/63S1681  
53/63S1681A



53/63S3281  
53/63S3281A



**Absolute Maximum Ratings**

Supply voltage $V_{CC}$ .....	-0.5V to 7V
Input voltage .....	-1.5V to 7V
Off-state output voltage .....	-0.5V to 5.5V
Storage temperature range .....	-65°C to +150°C

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
$V_{CC}$	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
$T_A$	Operating free-air temperature	-55		125	0		75	°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IL}$	Low-level input voltage				0.8		V
$V_{IH}$	High-level input voltage			2			V
$V_{IC}$	Input clamp voltage	$V_{CC} = \text{MIN}$	$I_I = -18\text{mA}$		-1.5		V
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$	$V_I = 0.4\text{V}$		-0.25		mA
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$	$V_I = V_{CC} \text{ MAX}$		40		$\mu\text{A}$
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	$I_{OL} = 16\text{mA}$	MIL		0.5	V
				COM except S1681, S3281		0.45	
				COM S1681, S3281		0.5	
$V_{OH}$	High-level output voltage*	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	MIL $I_{OH} = -2\text{mA}$	2.4			V
			COM $I_{OH} = -3.2\text{mA}$				
$I_{OZL}$	Off-state output current*	$V_{CC} = \text{MAX}$	$V_O = 0.4\text{V}$		-40	$\mu\text{A}$	
$I_{OZH}$			$V_O = 2.4\text{V}$		40	$\mu\text{A}$	
$I_{CEX}$	Open collector output current	$V_{CC} = \text{MAX}$	$V_O = 2.4\text{V}$		40	$\mu\text{A}$	
			$V_O = 5.5\text{V}$		100	$\mu\text{A}$	
$I_{OS}$	Output short-circuit current**	$V_{CC} = 5\text{V}$	$V_O = 0\text{V}$	-20		-90	mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$ All inputs grounded. All outputs open.	S080, S081		90	125	mA
			S140, S141		80	130	
			S240, S241		90	130	
			S441, S441A		95	140	
			S841, S841A		110	150	
			S1641, S1641A		130	175	
			S1681, S1681A		135	185	
S3281, S3281A		150	190				

\* Three-state only.

\*\* Not more than one output should be shorted at a time and duration of the short-circuit should not exceed one second.

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**Monolithic Memories** 



## Switching Characteristics

Over Commercial Operating Conditions

DEVICE TYPE	$t_{AA}$ (ns)	$t_{EA}$ AND $t_{ER}$ (ns)
	ADDRESS ACCESS TIME MAX	ENABLE ACCESS AND RECOVERY TIME MAX
63S080, 63S081	25	20
63S140, 63S141	45	25
63S240, 63S241	45	25
63S441	45	25
63S441A	35	25
63S841	50	25
63S841A	35	25
63S1641	50	25
63S1641A	35	25
63S1681	50	30
63S1681A	35	30
63S3281	50	30
63S3281A	40	30

Over Military Operating Conditions

DEVICE TYPE	$t_{AA}$ (ns)	$t_{EA}$ AND $t_{ER}$ (ns)
	ADDRESS ACCESS TIME MAX	ENABLE ACCESS AND RECOVERY TIME MAX
53S080, 53S081	35	30
53S140, 53S141	55	30
53S240, 53S241	55	30
53S441	55	30
53S441A	50	30
53S841	55	30
53S841A	50	30
53S1641	65	30
53S1641A	50	30
53S1681	60	35
53S1681A	50	35
53S3281	60	35
53S3281A	50	35

Monolithic Memories

MEMORY

# High Performance Registered 1024x4 PROM 53/63RA441

## Features/Benefits

- Edge triggered "D" registers
- Advanced Schottky processing
- 4-bit-wide in 18 pin for high board density
- Lower system package counts
- Lower system power
- Faster cycle times
- 16mA I<sub>OL</sub> output drive capability

## Applications

- Pipelined microprogramming
- State sequencers
- Next address generation
- Mapping PROM

## Description

A family of registered PROMs offers new savings for designers of pipelined microprogrammable systems. The wide instruction register which holds the micro-instruction during execution, is now incorporated into the PROM chip.

## Ordering Information

MEMORY		PACKAGE		DEVICE TYPE	
SIZE	ORGANIZATION	PINS	TYPE	MIL	COM
4K	1024x4	18	J, N	53RA441	63RA441

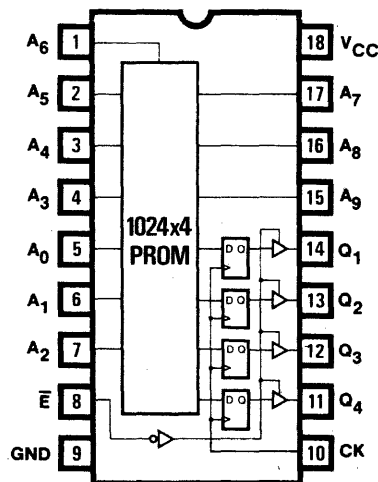
## Edge Triggered Register

The PROM output is loaded into a 4-bit register on the rising edge of the clock. The use of the term "register" is to be distinguished from the term "latch," in that a register contains master slave flip-flops and the latch contains gated flip-flops. The advantages of using a register are that system timing is simplified, and faster micro cycle times can be obtained.

The output of the register is buffered by three-state drivers which are compatible with the new low-power Schottky three-state bus standard.

## Pin Configuration

53/63RA441



**Absolute Maximum Ratings**

Supply voltage, $V_{CC}$ .....	7V
Input voltage .....	7V
Off-state output voltage .....	5.5V
Storage temperature .....	-65° to +150°C

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{CC}$	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
$t_{su}$	Address set-up time	60	30		50	30		
$t_h$	Address hold time	0	-10		0	-10		
$t_w$	Clock pulse width	25	8		20	8		
$T_A$	Operating free-air temperature	-55		125	0		75	°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IL}$	Low-level input voltage					0.8	V
$V_{IH}$	High-level input voltage			2			V
$V_{IC}$	Input clamp voltage	$V_{CC} = \text{MIN}$	$I_I = -18\text{mA}$			-1.5	V
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$	$V_I = 0.4\text{V}$			-0.25	mA
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$	$V_I = V_{CC}$			40	μA
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	$I_{OL} = 16\text{mA}$			0.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	MIL $I_{OH} = -2\text{mA}$ COM $I_{OH} = -3.2\text{mA}$	2.4			V
$I_{OZL}$	Off-state output current	$V_{CC} = \text{MAX}$	$V_O = 0.5\text{V}$			-40	μA
$I_{OZH}$			$V_O = 2.4\text{V}$			40	μA
$I_{OS}$	Output short-circuit current*	$V_{CC} = 5\text{V}$	$V_O = 0\text{V}$	-20		-90	mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$	All inputs grounded All outputs open	MIL COM	120 120	175 165	mA

\*Not more than one output should be shorted at a time and duration of the short-circuit should not exceed one second.

†Typicals at 5.0V  $V_{CC}$  and 25°C  $T_A$

**Switching Characteristics**

Over Operating Conditions

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP†	MAX	MIN	TYP†	MAX	
$t_{pd}$	Clock to output access time		20	35		20	30	ns
$t_{ER}/t_{EA}$	Enable to output access and recovery time		19	35		19	30	ns

# High Performance 1024x8 Registered PROM

## 53/63RS881 53/63RS881A

### Features/Benefits

- Edge triggered "D" registers
- Synchronous and Asynchronous enables
- Versatile 1:16 initialization words
- 8-bit-wide in 24 pin SKINNYDIP® for high board density
- Simplifies system timing
- Faster cycle times
- 16mA I<sub>OL</sub> output drive capability
- Reliable titanium-tungsten fuses (Ti-W)

### Applications

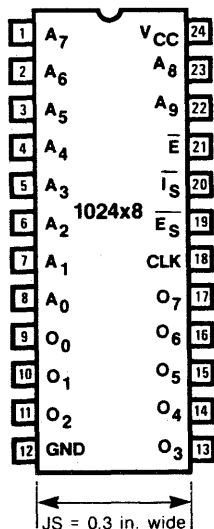
- Microprogram control store
- State sequencers
- Next address generation
- Mapping PROM

### Description

The 53/63RS881 and 53/63RS881A are 1Kx8 PROMs with on chip "D" type registers, versatile output enable control through synchronous and asynchronous enable inputs, and flexible start up sequencing through programmable initialization.

Data is transferred into the output registers on the rising edge of the clock. Provided that the asynchronous ( $\bar{E}$ ) and synchronous ( $\bar{E}_S$ ) enables are low, the data will appear at the outputs. Prior to the positive clock edge, register data are not

### Pin Configuration



### Ordering Information

MEMORY		PACKAGE		DEVICE TYPE	
SIZE	PERFORMANCE	PINS	TYPE	MIL	COM
8K	Standard	24	JS, F	53RS881	63RS881
	Enhanced	28	L	53RS881A	63RS881A

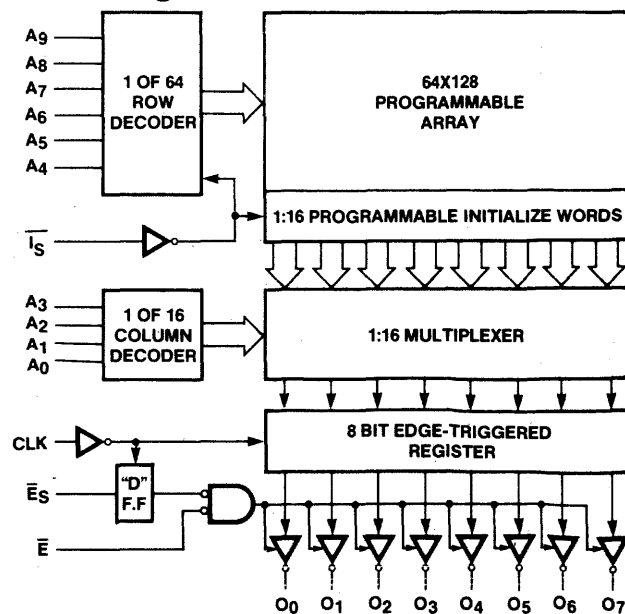
\* Flat-pack — contact the factory

affected by changes in addressing or synchronous enable inputs.

Memory expansion and data control is made flexible with synchronous and asynchronous enable inputs. Outputs may be set to the high impedance state at any time by setting  $\bar{E}$  to a high or if  $\bar{E}_S$  is high when the rising clock edge occurs. When  $V_{CC}$  power is first applied the synchronous enable flip-flop will be in the set condition causing the outputs to be in the high impedance state.

The flexible initialization feature allows start up and time out sequencing with 1:16 programmable words to be loaded into the output registers. With the synchronous INITIALIZE ( $\bar{I}_S$ ) pin low, one of the 16 column words ( $A_3-A_0$ ) will be set in the output registers independent of the row addresses ( $A_9-A_4$ ). The unprogrammed state of  $\bar{I}_S$  words are low, presenting a CLEAR with  $\bar{I}_S$  pin low. With all  $\bar{I}_S$  column words ( $A_3-A_0$ ) programmed to the same pattern, the  $\bar{I}_S$  function will be independent of both row and column addressing and may be used as a single pin control. With all  $\bar{I}_S$  words programmed high a PRESET function is performed.

### Block Diagram



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### Absolute Maximum Ratings

Supply voltage $V_{CC}$ .....	Operating -0.5 to 7V .....	Programming 12V .....
Input voltage .....	-1.5 to 7V .....	7V .....
Off-state output voltage .....	-0.5V to 5.5V .....	
Storage temperature .....	-65°C to +150°C .....	

### Operating Conditions

SYMBOL	PARAMETER	TYP	MILITARY				COMMERCIAL				UNIT
			53RS881A		53RS881		63RS881A		63RS881		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
$t_w$	Width of clock (high or low)	10	20		20		20		20		ns
$t_{s(A)}$	Setup time from address to clock	25	40		45		30		35		ns
$t_{s(\overline{E_S})}$	Setup time from $\overline{E_S}$ to clock	8	15		15		15		15		ns
$t_{s(\overline{I_S})}$	Setup time from $\overline{I_S}$ to clock	20	30		35		25		30		ns
$t_{h(A)}$	Hold time address to clock	-5	0		0		0		0		ns
$t_{h(\overline{E_S})}$	Hold time ( $\overline{E_S}$ )	-3	5		5		5		5		ns
$t_{h(\overline{I_S})}$	Hold time ( $\overline{I_S}$ )	-5	0		0		0		0		ns
$V_{CC}$	Supply voltage	5	4.5	5.5	4.5	5.5	4.75	5.25	4.75	5.25	V
$T_A$	Operating free-air temperature	25	-55	125	-55	125	0	75	0	75	°C

### Electrical Characteristics Over Operating Conditions

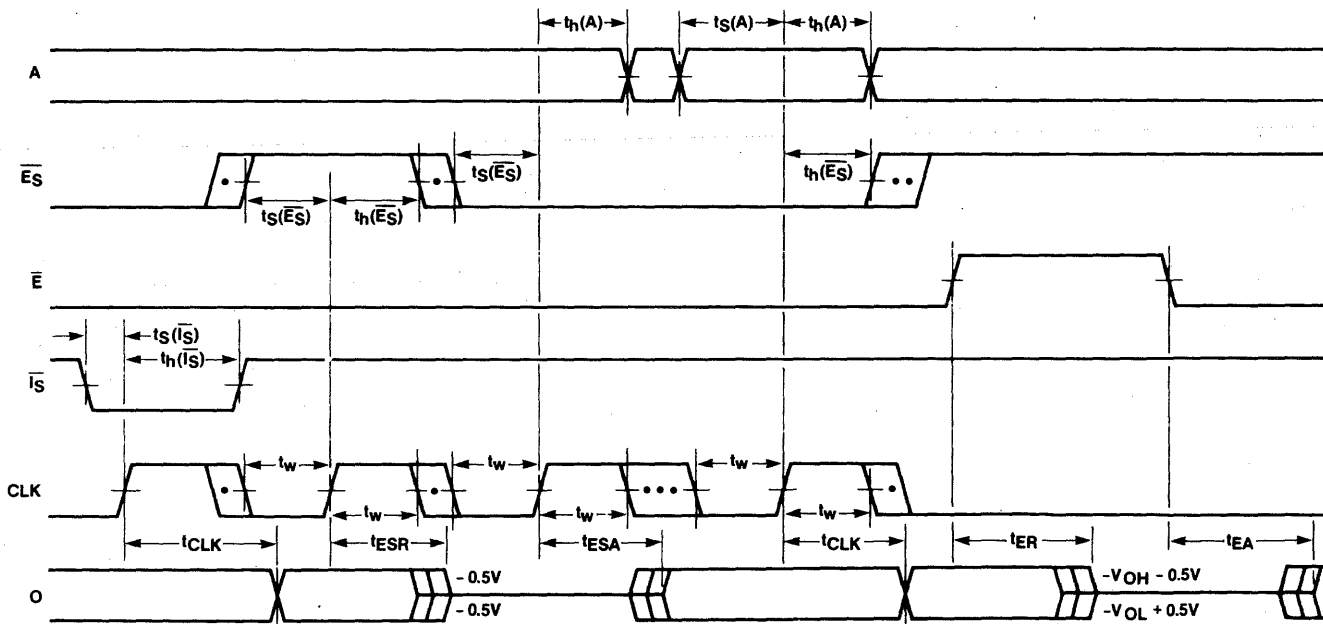
SYMBOL	PARAMETER	TEST CONDITIONS	t			UNIT	
			MIN	TYP	MAX		
$V_{IL}$	Low-level input voltage				0.8	V	
$V_{IH}$	High-level input voltage				2	V	
$V_{IC}$	Input clamp voltage	$V_{CC} = \text{MIN}$ $I_I = -18\text{mA}$			-1.2	V	
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$ $V_I = 0.4\text{V}$			-0.25	mA	
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$ $V_I = V_{CC}\text{MAX}$			40	$\mu\text{A}$	
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$ $I_{OL} = 16\text{mA}$			0.5	V	
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$ MIL $I_{OH} = -2\text{mA}$ COM $I_{OH} = -3.2\text{mA}$			2.4	V	
$I_{OZL}$	Off-state output current	$V_{CC} = \text{MAX}$	$V_O = 0.4\text{V}$		-40	$\mu\text{A}$	
$I_{OZH}$			$V_O = 2.4\text{V}$		40		
$I_{OS}$	Output short-circuit current*	$V_{CC} = \text{MAX}$	$V_O = 0\text{V}$		-20	-90	mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$	All inputs TTL; all outputs open.		130	180	mA

\* Not more than one output should be shorted at a time and duration of the short-circuit should not exceed one second.  
 † Typical at 5.0V,  $V_{CC}$  and 25°C TA.

**Switching Characteristics** Over Operating Conditions and using Standard Test Load

SYMBOL	PARAMETER	TYP	MILITARY		COMMERCIAL				UNIT		
			53RS881A		53RS881		63RS881A			63RS881	
			MIN	MAX	MIN	MAX	MIN	MAX		MIN	MAX
$t_{CLK}$	Clock to output Delay	10	20	25	15	20			ns		
$t_{ESA}$	Clock to output access time ( $\overline{E_S}$ )	18	30	35	25	30			ns		
$t_{ESR}$	Clock to output recovery time ( $\overline{E_S}$ )	17	30	35	25	30			ns		
$t_{EA}$	Enable to output access time ( $\overline{E}$ )	18	30	35	25	30			ns		
$t_{ER}$	Disable to output recovery time ( $\overline{E}$ )	17	30	35	25	30			ns		

**Definition of Waveforms**



- NOTES: 1. Input pulse amplitude 0V to 3.0V.  
 2. Input rise and fall times 2-5ns from 1.0V to 2.0V.  
 3. Input access measured at the 1.5V level.  
 4.  $t_{AA}$  is tested with switch  $S_1$  closed.  $C_L = 30pF$  and measured at 1.5V output level.  
 5.  $t_{EA}$  and  $t_{ESA}$  are measured at the 1.5V output level with  $C_L = 30pF$ .  $S_1$  is open for high impedance to "1" test and closed for high impedance to "0" test.  
 $t_{ER}$  and  $t_{ESR}$  are measured  $C_L = 5pF$ .  $S_1$  is open for "1" to high impedance test, measured at  $V_{OH} - 0.5V$  output level;  $S_1$  is closed for "0" to high impedance test measured at  $V_{OL} + 0.5V$  output level.

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# 2048x8 Registered Prom with Asynchronous Enable

# 53/63RA1681 53/63RA1681A

## Features/Benefits

- Asynchronous output enable
- Edge-triggered "D" registers
- Versatile 1:16 user programmable initialization words
- 8-bit-wide in 24 pin SKINNYDIP® for high board density
- Simplifies system timing
- Faster cycle times
- 16mA  $I_{OL}$  output drive capability
- Reliable titanium-tungsten fuses (Ti-W)

## Applications

- Microprogram control store
- State sequencers
- Next address generation
- Mapping PROM

## Description

The 53/63RS1681 and 53/63RS1681A are 2 K x 8 PROMs with on chip "D" type registers. Output enable control through an asynchronous enable input and flexible start up sequencing through programmable initialization words.

## Ordering Information

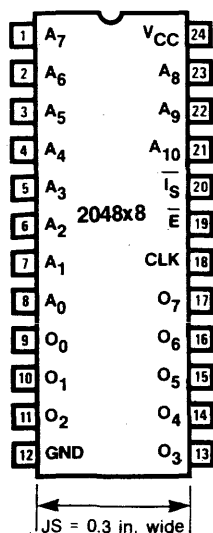
MEMORY		PACKAGE		DEVICE TYPE	
SIZE	PERFORMANCE	PINS	TYPE	MIL	COM
16K	Standard	24	JS	53RA1681	63RA1681
	Enhanced			53RA1681A	63RA1681A

Data is transferred into the output registers on the rising edge of the clock. Provided that the asynchronous ( $\bar{E}$ ) enable is LOW, the data will appear at the outputs. Prior to the positive clock edge, register data are not affected by changes in addressing.

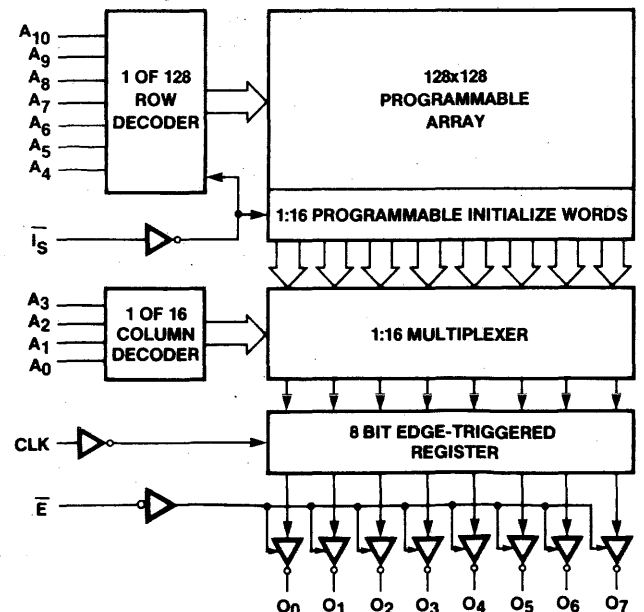
Memory expansion and data control is made flexible with asynchronous enable input. Outputs may be set to the high impedance state at any time by setting  $\bar{E}$  to a HIGH.

The flexible initialization feature allows start up and time out sequencing with 1:16 programmable words to be loaded into the output registers. With the synchronous INITIALIZE ( $\bar{I}_S$ ) pin LOW, one of the 16 column words ( $A_3-A_0$ ) will be set in the output registers independent of the row addresses ( $A_{10}-A_4$ ). With all  $\bar{I}_S$  column words ( $A_3-A_0$ ) programmed to the same pattern, the  $\bar{I}_S$  function will be independent of both row and column addressing and may be used as a single pin control. With all  $\bar{I}_S$  words programmed HIGH a PRESET function is performed. The unprogrammed state of  $\bar{I}_S$  words are LOW, presenting a CLEAR with  $\bar{I}_S$  pin LOW.

## Pin Configuration



## Block Diagram



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**Absolute Maximum Ratings**

Supply voltage $V_{CC}$ .....	Operating -0.5 to 7V .....	Programming 12V
Input voltage .....	-1.5 to 7V .....	7V
Off-state output voltage .....	-0.5V to 5.5V .....	12V
Storage temperature .....	-65°C to +150°C	

**Operating Conditions**

SYMBOL	PARAMETER	TYP†	MILITARY				COMMERCIAL				UNIT
			53RA1681A		53RA1681		63RA1681A		63RA1681		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
$t_w$	Width of clock (high or low)	10	20		20		20		20		ns
$t_{s(A)}$	Setup time from address to clock	28	40		45		35		40		ns
$t_{s(\overline{IS})}$	Setup time from $\overline{IS}$ to clock	20	30		35		25		30		ns
$t_{h(A)}$	Hold time address to clock	-5	0		0		0		0		ns
$t_{h(\overline{IS})}$	Hold time ( $\overline{IS}$ )	-5	0		0		0		0		ns
$V_{CC}$	Supply voltage	5	4.5	5.5	4.5	5.5	4.75	5.25	4.75	5.25	V
$T_A$	Operating free-air temperature	25	-55	125	-55	125	0	75	0	75	°C

**Electrical Characteristics** Over Operating Conditions

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IL}$	Low-level input voltage					0.8	V
$V_{IH}$	High-level input voltage			2.0			V
$V_{IC}$	Input clamp voltage	$V_{CC} = \text{MIN}$	$I_I = -18\text{mA}$			-1.2	V
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$	$V_I = 0.4\text{V}$			-0.25	mA
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$	$V_I = V_{CC}\text{MAX}$			40	μA
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	$I_{OL} = 16\text{mA}$			0.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	MIL $I_{OH} = -2\text{mA}$ COM $I_{OH} = -3.2\text{mA}$	2.4			V
$I_{OZL}$	Off-state output current	$V_{CC} = \text{MAX}$	$V_O = 0.4\text{V}$			-40	μA
$I_{OZH}$			$V_O = 2.4\text{V}$			40	
$I_{OS}$	Output short-circuit current*	$V_{CC} = 5.0\text{V}$	$V_O = 0\text{V}$	-20		-90	mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$	All inputs TTL; all outputs open.	140		185	mA

\* Not more than one output should be shorted at a time and duration of the short-circuit should not exceed one second.

† Typical at 5.0 V  $V_{CC}$  and 25°C  $T_A$ .

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**Switching Characteristics Over Operating Conditions and using Standard Test Load**

SYMBOL	PARAMETER	TYP <sup>†</sup>	MILITARY				COMMERCIAL				UNIT
			53RA1681A		53RA1681		63RA1681A		63RA1681		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
t <sub>CLK</sub>	Clock to output Delay	10	20		25		15		20		ns
t <sub>EA</sub>	Enable to output access time (E)	15	30		35		25		30		ns
t <sub>ER</sub>	Disable to output recovery time (E)	15	30		35		25		30		ns

<sup>†</sup> Typical at 5.0 V V<sub>CC</sub> and 25°C T<sub>A</sub>.

**Definition of Waveforms**



NOTES: 1. Input pulse amplitude 0V to 3.0V.

2. Input rise and fall times 2-5ns from 1.0V to 2.0V.

3. Input access measured at the 1.5V level.

4. Switch S<sub>1</sub> is closed, C<sub>L</sub> = 30pF and outputs measured at 1.5V level for all tests except t<sub>EA</sub> and t<sub>ER</sub>.

5. t<sub>EA</sub> is measured at the 1.5V output level with C<sub>L</sub> = 30 pF. S<sub>1</sub> is open for high impedance to "1" test and closed for high impedance to "0" test.

t<sub>ER</sub> is tested with C<sub>L</sub> = 5pF. S<sub>1</sub> is open for "1" to high impedance test, measured at V<sub>OH</sub> -0.5V output level; S<sub>1</sub> is closed for "0" to high impedance test measured at V<sub>OL</sub> +0.5V output level.

# 2048x8 Registered Prom with Synchronous Enable

# 53/63RS1681 53/63RS1681A

## Features/Benefits

- Synchronous output enable
- Edge-triggered "D" registers
- Versatile 1:16 user programmable initialization words
- 8-bit-wide in 24 pin SKINNYDIP® for high board density
- Simplifies system timing
- Faster cycle times
- 16mA I<sub>OL</sub> output drive capability
- Reliable titanium-tungsten fuses (Ti-W)

## Applications

- Microprogram control store
- State sequencers
- Next address generation
- Mapping PROM

## Description

The 53/63RS1681 and 53/63RS1681A are 2 K x 8 PROMs with on chip "D" type registers, versatile output enable control through synchronous enable input and flexible start up sequencing through programmable initialization words.

Data is transferred into the output registers on the rising edge of the clock. Provided that the synchronous ( $\overline{E_S}$ ) enable is LOW, the data will appear at the outputs. Prior to the positive

## Ordering Information

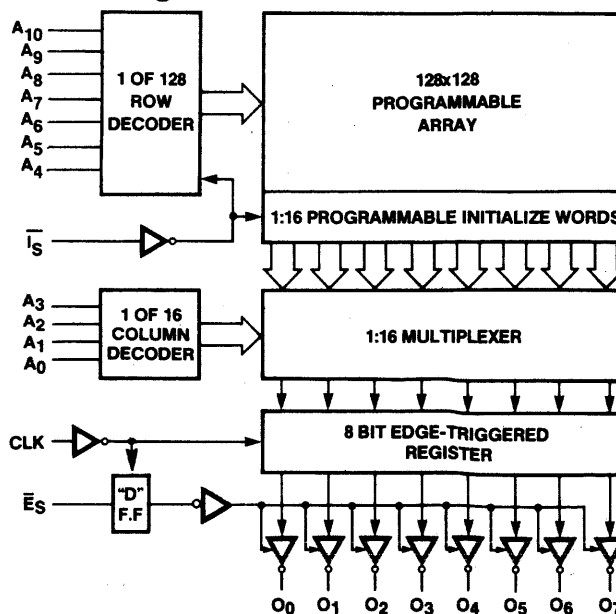
MEMORY		PACKAGE		DEVICE TYPE	
SIZE	PERFORMANCE	PINS	TYPE	MIL	COM
16K	Standard	24	JS	53RS1681	63RS1681
	Enhanced			53RS1681A	63RS1681A

clock edge, register data are not affected by changes in addressing or synchronous enable inputs.

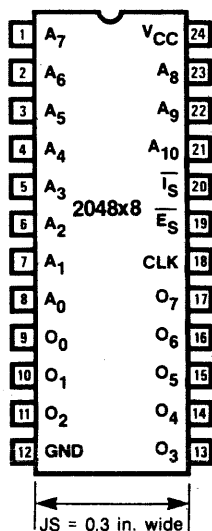
Memory expansion and data control is made flexible with the synchronous enable input. Outputs may be set to the high impedance state by setting  $\overline{E_S}$  HIGH before the rising clock edge occurs. When  $V_{CC}$  power is first applied the synchronous enable flip-flop will be in the set condition causing the outputs to be in the high impedance state.

The flexible initialization feature allows start up and time out sequencing with 1:16 programmable words to be loaded into the output registers. With the synchronous INITIALIZE ( $\overline{I_S}$ ) pin LOW, one of the 16 column words ( $A_3-A_0$ ) will be set in the output registers independent of the row addresses ( $A_{10}-A_4$ ). With all  $\overline{I_S}$  column words ( $A_3-A_0$ ) programmed to the same pattern, the  $\overline{I_S}$  function will be independent of both row and column addressing and may be used as a single pin control. With all  $\overline{I_S}$  words programmed HIGH a PRESET function is performed. The unprogrammed state of  $\overline{I_S}$  words are LOW, presenting a CLEAR with  $\overline{I_S}$  pin LOW.

## Block Diagram



## Pin Configuration



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**Absolute Maximum Ratings**

Supply voltage $V_{CC}$ .....	Operating .....	7V	Programming .....	12V
Input voltage .....	.....	-1.5 to 7V	.....	7V
Off-state output voltage .....	.....	-0.5V to 5.5V	.....	12V
Storage temperature .....	.....	-65°C to +150°C		

**Operating Conditions**

SYMBOL	PARAMETER	TYP†	MILITARY				COMMERCIAL				UNIT
			53RS1681A		53RS1681		63RS1681A		63RS1681		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
$t_w$	Width of clock (high or low)	10	20		20		20		20		ns
$t_{s(A)}$	Setup time from address to clock	28	40		45		35		40		ns
$t_{s(\overline{E_S})}$	Setup time from $\overline{E_S}$ to clock	7	15		15		15		15		ns
$t_{s(\overline{I_S})}$	Setup time from $\overline{I_S}$ to clock	20	30		35		25		30		ns
$t_{h(A)}$	Hold time address to clock	-5	0		0		0		0		ns
$t_{h(\overline{E_S})}$	Hold time ( $\overline{E_S}$ )	-3	5		5		5		5		ns
$t_{h(\overline{I_S})}$	Hold time ( $\overline{I_S}$ )	-5	0		0		0		0		ns
$V_{CC}$	Supply voltage	5	4.5	5.5	4.5	5.5	4.75	5.25	4.75	5.25	V
$T_A$	Operating free-air temperature	25	-55	125	-55	125	0	75	0	75	°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IL}$	Low-level input voltage					0.8	V
$V_{IH}$	High-level input voltage			2.0			V
$V_{IC}$	Input clamp voltage	$V_{CC} = \text{MIN}$	$I_I = -18\text{mA}$			-1.2	V
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$	$V_I = 0.4\text{V}$			-0.25	mA
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$	$V_I = V_{CC}$			40	μA
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	$I_{OL} = 16\text{mA}$			0.5	V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	MIL $I_{OH} = -2\text{mA}$ COM $I_{OH} = -3.2\text{mA}$	2.4			V
$I_{OZL}$	Off-state output current	$V_{CC} = \text{MAX}$	$V_O = 0.4\text{V}$			-40	μA
$I_{OZH}$			$V_O = 2.4\text{V}$			40	
$I_{OS}$	Output short-circuit current*	$V_{CC} = 5\text{V}$	$V_O = 0\text{V}$	-20		-90	mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$	All inputs TTL; all outputs open.		140	185	mA

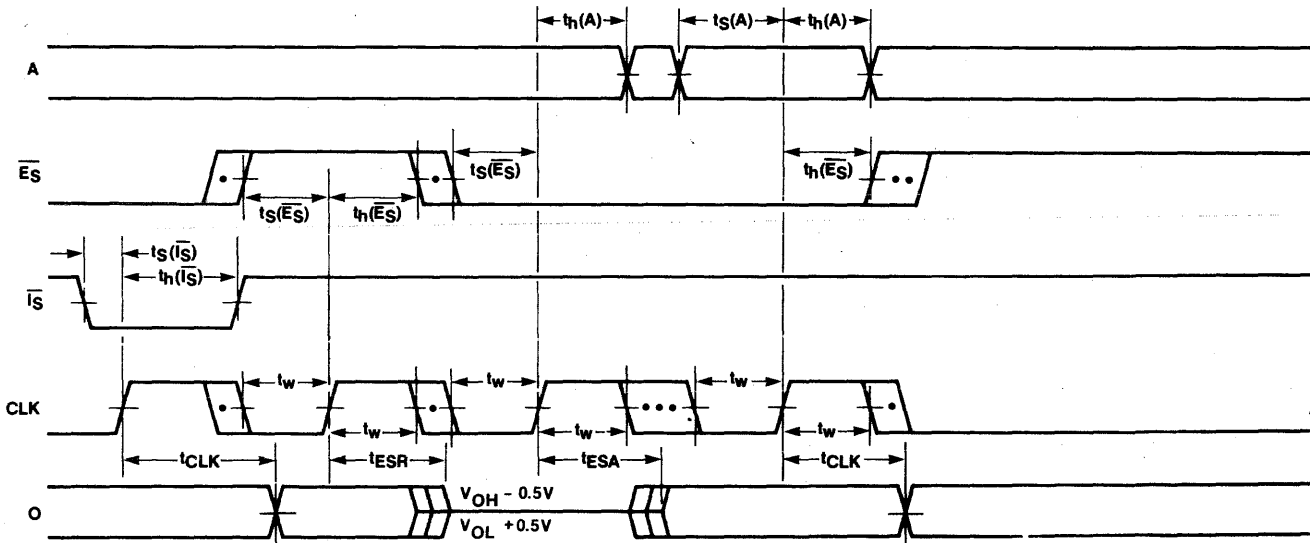
\* Not more than one output should be shorted at a time and duration of the short-circuit should not exceed one second.  
† Typical at 5.0 V  $V_{CC}$  and 25°C  $T_A$

**Switching Characteristics** Over Operating Conditions and using Standard Test Load

SYMBOL	PARAMETER	TYP <sup>†</sup>	MILITARY				COMMERCIAL				UNIT
			53RS1681A		53RS1681		63RS1681A		63RS1681		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
t <sub>CLK</sub>	Clock to output Delay	10	20		25		15		20	ns	
t <sub>ESA</sub>	Clock to output access time ( $\overline{E_S}$ )	15	30		35		25		30	ns	
t <sub>ESR</sub>	Clock to output recovery time ( $\overline{E_S}$ )	15	30		35		25		30	ns	

<sup>†</sup> Typical at 5.0 V V<sub>CC</sub> and 25°C T<sub>A</sub>.

**Definition of Waveforms**



- NOTES:
1. Input pulse amplitude 0V to 3.0V.
  2. Input rise and fall times 2-5ns from 1.0V to 2.0V.
  3. Input access measured at the 1.5V level.
  4. Switch S<sub>1</sub> is closed, C<sub>L</sub> = 30pF and outputs measured at 1.5V level for all tests except t<sub>ESA</sub> and t<sub>ESR</sub>.
  5. t<sub>ESA</sub> is measured at the 1.5V output level with C<sub>L</sub> = 30pF. S<sub>1</sub> is open for high impedance to "1" test and closed for high impedance to "0" test.  
 t<sub>ESR</sub> is tested with C<sub>L</sub> = 5pF. S<sub>1</sub> is open for "1" to high impedance test, measured at V<sub>OH</sub>-0.5V output level; S<sub>1</sub> is closed for "0" to high impedance test measured at V<sub>OL</sub>+0.5V output level.

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# 4096x4 Diagnostic Registered PROM with Asynchronous Enable

# 53D1641 63D1641

Patent Pend.

## Features/Benefits

- Asynchronous output enable
- Provides system diagnostic testing for system controllability and observability
- Shadow register eliminates shifting hazards
- Edge-triggered "D" registers simplifies system timing
- Casadable for wide control words used in microprogramming
- 24-pin SKINNYDIP® saves space
- 24 mA output drive capability
- Replaces embedded diagnostic code

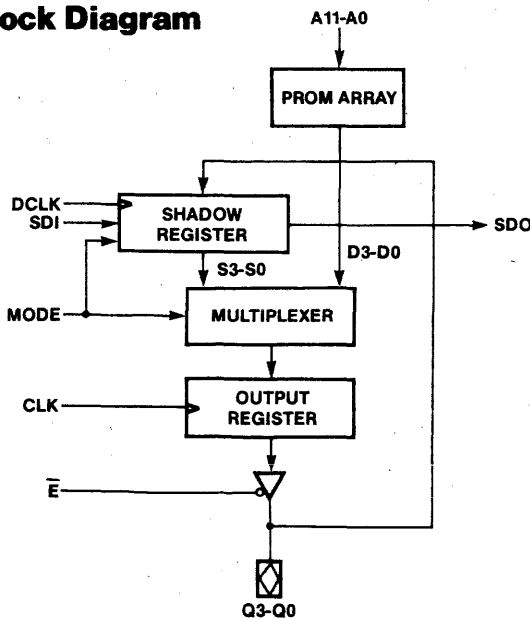
## Applications

- Microprogram control store with built-in system diagnostic testing
- Serial character generator
- Serial code converter
- Parallel in/serial out memory
- Cost-effective board testing

## Description

The 53/63D1641 is a 4Kx4 PROM with registered three-state outputs and a shadow register for diagnostic capabilities.

## Block Diagram



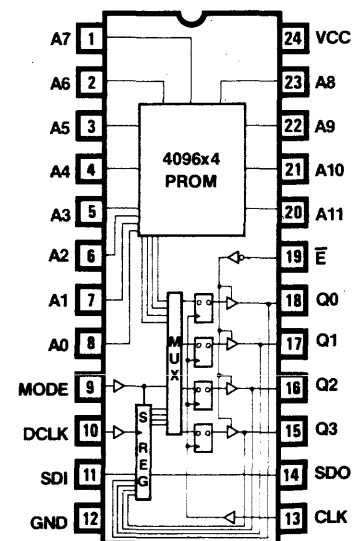
SKINNYDIP® is a registered trademark of Monolithic Memories.

## Ordering Information

MEMORY		TEMP.	PACKAGE		PART NO.
SIZE	ORG.		PINS	TYPE	
16K	4096x4	MIL	24	JS	53D1641
		COM			63D1641

Shadow register diagnostics allow observation and control of the system without introducing intermediate illegal states. The output register, which can receive parallel data from either the PROM array or the shadow register, is loaded on the rising edge of CLK. The shadow register, which can receive parallel data from the output register or serial data from SDI, is loaded on the rising edge of DCLK. When the output drivers are disabled, the shadow register receives its parallel data from the output bus. During diagnostics, data loaded into the output register from the PROM array can be parallel-loaded into the shadow register and serially shifted out through SDO, allowing observation of the system. Similarly, diagnostic data can be serially shifted into the shadow register through SDI, and parallel-loaded into the output register, allowing control and test scanning to be imposed on the system. Since the output register and the shadow register are loaded by different input signals, they can be operated independent of one another. In addition, diagnostic PROMs can be cascaded to construct wide control words used in microprogramming.

## Logic Symbol



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**Monolithic Memories**

## Function Table

INPUTS				OUTPUTS			OPERATION
MODE	SDI	CLK	DCLK	Q <sub>3</sub> -Q <sub>0</sub>	S <sub>3</sub> -S <sub>0</sub>	SDO	
L	X	↑	*	Q <sub>n</sub> ← PROM	HOLD	S <sub>3</sub>	Load output register from PROM array
L	X	*	↑	HOLD	S <sub>n</sub> ← S <sub>n-1</sub> S <sub>0</sub> ← SDI	S <sub>3</sub>	Shift shadow register data
L	X	↑	↑	Q <sub>n</sub> ← PROM	S <sub>n</sub> ← S <sub>n-1</sub> S <sub>0</sub> ← SDI	S <sub>3</sub>	Load output register from PROM array while shifting shadow register data
H	X	↑	*	Q <sub>n</sub> ← S <sub>n</sub>	HOLD	SDI	Load output register from shadow register
H	L	*	↑	HOLD	S <sub>n</sub> ← Q <sub>n</sub>	SDI	Load shadow register from output bus
H	H	*	↑	HOLD	HOLD	SDI	No operation †

\* Clock must be steady or falling.

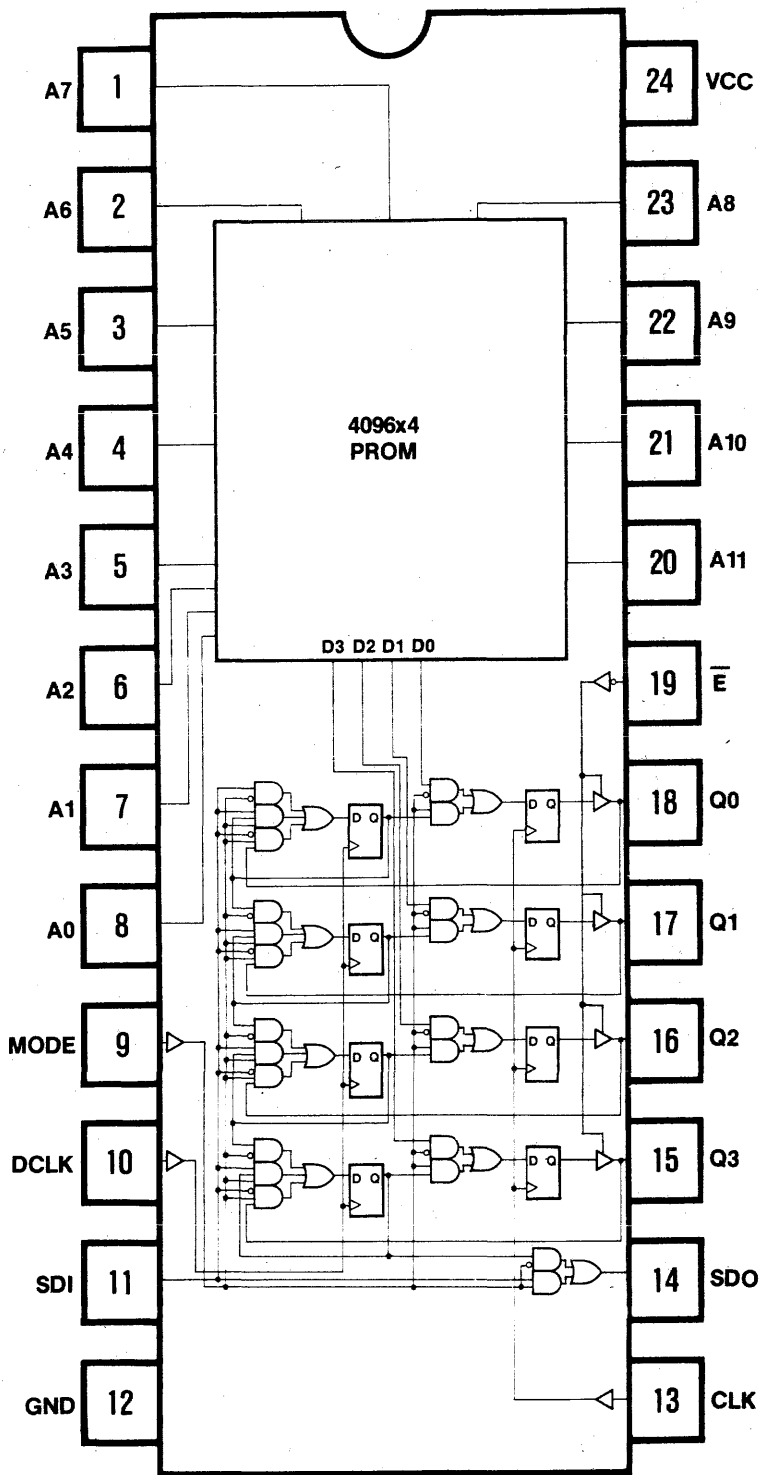
† Reserved operation for SN54/74S818 8-Bit Diagnostic Register.

## Definition of Signals

MODE	The MODE pin controls the output register multiplexer and the shadow register. When MODE is LOW, the output register receives data from the PROM array and the shadow register is configured as a shift register with SDI as its input. When MODE is HIGH, the output register receives data from the shadow register. The shadow register is controlled by SDI as well as MODE. With MODE HIGH and SDI LOW, the shadow register receives parallel data from the output bus. With MODE and SDI both HIGH, the shadow register holds its present data.	CLK	The clock pin loads the output register on the rising edge of CLK.
SDI	The Serial Data In pin is the input to the least significant bit of the shadow register when operating in the shift mode. SDI is also a control input to the shadow register when it is not in the shift mode.	DCLK	The diagnostic clock pin loads or shifts the shadow register on the rising edge of DCLK.
SDO	The Serial Data Out pin is the output from the most significant bit of the shadow register when operating in the shift mode. When the shadow register is not in the shift mode, SDO displays the logic level present at SDI, decreasing serial shift time for cascaded diagnostic PROMs.	Q <sub>3</sub> -Q <sub>0</sub>	Q <sub>n</sub> represents the data outputs of the output register. During a shadow register load with outputs enabled these pins are the internal data inputs to the shadow register. With the outputs three-stated these pins are external data inputs to the shadow register.
		S <sub>3</sub> -S <sub>0</sub>	S <sub>n</sub> represents the internal shadow register outputs.
		A <sub>11</sub> -A <sub>0</sub>	A <sub>n</sub> represents the address inputs to the PROM array.
		$\bar{E}$	The Output Enable pin operates independent of CLK. When $\bar{E}$ is LOW the outputs are enabled. When $\bar{E}$ is HIGH, the outputs are in the high impedance state.

Logic Diagram

**4096x4 Diagnostic PROM  
with Asynchronous Enable**



Monolithic Memories

MEMORY

**Absolute Maximum Ratings**

	Operating	Programming
Supply voltage $V_{CC}$ .....	-0.5V to 7V .....	12V .....
Input voltage .....	-1.5V to 7V .....	7V .....
Input current .....	-30mA to +5mA .....	
Off-state output voltage .....	-0.5V to 5.5V .....	12V .....
Storage temperature .....	-65°C to +150°C .....	

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP†	MAX	MIN	TYP†	MAX	
$V_{CC}$	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
$T_A$	Operating free air temperature	-55	25	125	0	25	75	°C
$t_w$	Width of CLK (HIGH or LOW)	25	10		20	10		ns
$t_{su}$	Set up time from address to CLK	45	25		40	25		ns
$t_h$	Hold time for CLK	0	-15		0	-15		ns
$t_{wd}$	Width of DCLK (HIGH or LOW)	45	15		40	15		ns
$t_{sud}$	Set up time from control inputs (SDI, MODE) to CLK, DCLK	50	20		45	20		ns
$t_{hd}$	Hold time for DCLK	0	-5		0	-5		ns

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
$V_{IL}$	Low-level input voltage				0.8		V
$V_{IH}$	High-level input voltage			2			V
$V_{IC}$	Input clamp voltage	$V_{CC} = \text{MIN}$	$I_I = -18\text{mA}$		-1.2		V
$I_{IL}$	Low-level input current	$V_{CC} = \text{MAX}$	$V_I = 0.4\text{V}$		-0.25		mA
$I_{IH}$	High-level input current	$V_{CC} = \text{MAX}$	$V_I = V_{CC}$		40		uA
$V_{OL}$	Low-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	MIL $I_{OL} = 16\text{mA}$ COM $I_{OL} = 24\text{mA}$		0.5		V
$V_{OH}$	High-level output voltage	$V_{CC} = \text{MIN}$ $V_{IL} = 0.8\text{V}$ $V_{IH} = 2\text{V}$	MIL $I_{OH} = -2\text{mA}$ COM $I_{OH} = -3.2\text{mA}$	2.4			V
$I_{OZL}$	Off-state output current	$V_{CC} = \text{MAX}$	$V_O = 0.4\text{V}$		-100		uA
$I_{OZH}$			$V_O = 2.4\text{V}$		40		uA
$I_{OS}$	Output short-circuit current*	$V_{CC} = \text{MAX}$	$V_O = 0\text{V}$	-20	-90		mA
$I_{CC}$	Supply current	$V_{CC} = \text{MAX}$ , All inputs TTL; All outputs open		140	190		mA

† Typical at 5.0V  $V_{CC}$  and 25°C  $T_A$ .

\* Not more than one output should be shorted at a time and duration of the short-circuit should not exceed one second.

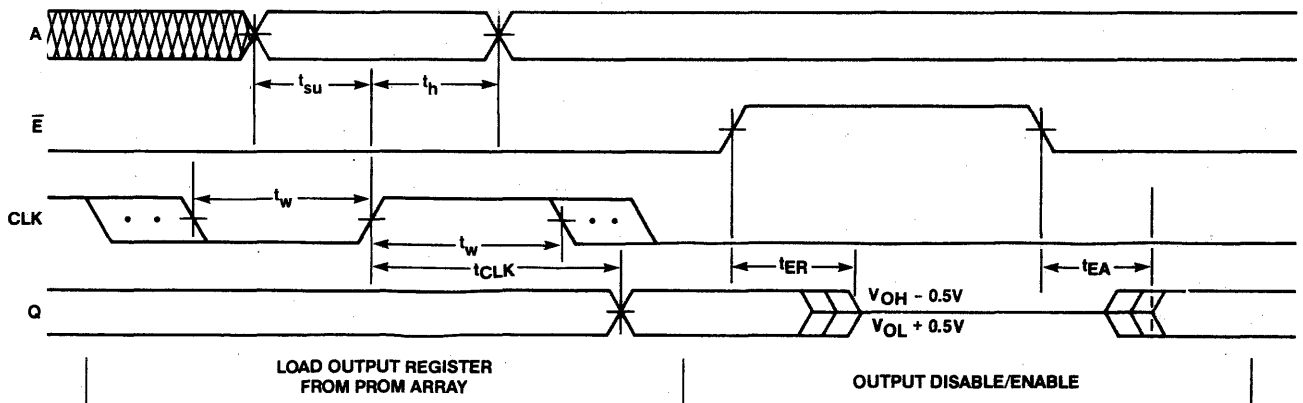


**Switching Characteristics Over Operating Conditions and Using Standard Test Load**

SYMBOL	PARAMETER	MILITARY		COMMERCIAL		UNIT	
		MIN	TYP†	MAX	MIN		TYP†
$t_{CLK}$	CLK to output		11	25	11	20	ns
$t_{ER}$	Disable time		16	30	16	25	ns
$t_{EA}$	Enable time		16	30	16	25	ns
$f_{MAXD}$	Maximum diagnostic clock frequency	7	18		10	18	MHz
$t_{DS}$	DCLK to SDO delay (MODE = LOW)		17	35	17	30	ns
$t_{SS}$	SDI to SDO delay (MODE = HIGH)		16	30	16	25	ns
$t_{MS}$	MODE to SDO delay		14	30	14	25	ns

† Typical at 5.0V  $V_{CC}$  and 25°C  $T_A$ .

**Definition of Waveforms**



**NORMAL PROM OPERATION (MODE = LOW)**

Monolithic Memories

MEMORY



# MOTOROLA SEMICONDUCTORS

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

## Memory Selector Guide

Motorola has developed a very broad range of reliable MOS and bipolar memories for virtually any digital data processing system application. And for those whose requirements go beyond individual components, Motorola also supplies Memory Systems and Micromodules.

New Motorola memories are being introduced continually. **This selector guide lists all those available as of November 1983.** For later releases, additional technical information or pricing, contact your nearest authorized Motorola distributor or Motorola sales office.

## RAMs

### MOS DYNAMIC RAMs

Organization	Part Number	Access Time (ns Max)	Power Supplies	No. of Pins
16384 x 1	MCM4116BP15	150	+12, ±5 V	16
16384 x 1	MCM4116BP20	200	+12, ±5 V	16
16384 x 1	MCM4116BP25	250	+12, ±5 V	16
16384 x 1	MCM4517P10	100	+5 V	16
16384 x 1	MCM4517P12	120	+5 V	16
16384 x 1	MCM4517P15	150	+5 V	16
16384 x 1	MCM4517P20	200	+5 V	16
65536 x 1	MCM6664AP15 <sup>1</sup>	150	+5 V	16
65536 x 1	MCM6664AP20 <sup>1</sup>	200	+5 V	16
65536 x 1	MCM6665AP15	150	+5 V	16
65536 x 1	MCM6665AP20	200	+5 V	16
65536 x 1	MCM6664BP15 <sup>1*</sup>	150	+5 V	16
65536 x 1	MCM6664BP20 <sup>1*</sup>	200	+5 V	16
65536 x 1	MCM6665BP15*	150	+5 V	16
65536 x 1	MCM6665BP20*	200	+5 V	16

### CMOS STATIC RAMs (+5 Volts)

Organization	Part Number	Access Time (ns max)	No. of Pins
2048 x 8	MCM6116P12	120	24
2048 x 8	MCM6116P15	150	24
2048 x 8	MCM6116P20	200	24
4096 x 1	MCM6147P55	55	18
4096 x 1	MCM6147P70	70	18

Operating temperature ranges:

MOS - 0°C to 70°C

ECL - 0°C to 75°C

TTL - Military - 55°C to +125°C, Commercial 0°C to 75°C

\* To be introduced.

(Not all speed selections shown)

<sup>1</sup> Motorola's innovative pin #1 refresh

<sup>2</sup> 300 mil package

### MOS STATIC RAMs (+5 Volts)

Organization	Part Number	Access Time (ns max)	No. of Pins
128 x 8	MCM6810	450	24
128 x 8	MCM68A10	360	24
128 x 8	MCM68B10	250	24
1024 x 4	MCM2114P20	200	18
1024 x 4	MCM2114P25	250	18
1024 x 4	MCM2114P30	300	18
1024 x 4	MCM2114P45	450	18
1024 x 4	MCM21L14P20	200	18
1024 x 4	MCM21L14P25	250	18
1024 x 4	MCM21L14P30	300	18
1024 x 4	MCM21L14P45	450	18
2048 x 8	MCM2016HP45	45	24
2048 x 8	MCM2016HN45	45	24 <sup>2</sup>
2048 x 8	MCM2016HY45	45	24 <sup>2</sup>
2048 x 8	MCM2016HP55	55	24
2048 x 8	MCM2016HN55	55	24 <sup>2</sup>
2048 x 8	MCM2016HY55	55	24 <sup>2</sup>
2048 x 8	MCM2016HP70	70	24
2048 x 8	MCM2016HN70	70	24 <sup>2</sup>
2048 x 8	MCM2016HY70	70	24 <sup>2</sup>
16384 x 1	MCM2167HP35	35	20
16384 x 1	MCM2167HL35	35	20
16384 x 1	MCM2167HZ35	35	20
16384 x 1	MCM2167HP45	45	20
16384 x 1	MCM2167HL45	45	20
16384 x 1	MCM2167HZ45	45	20
16384 x 1	MCM2167HP70	70	20
16384 x 1	MCM2167HL70	70	20
16384 x 1	MCM2167HZ70	70	20

### TTL RAMs

Organization	Part Number	Access Time (ns max)	Output	No. of Pins
1024 x 1	MCM93415	45	O.C.	16
1024 x 1	MCM93425	45	3-State	16
256 x 4	MCM93L422	60	3-State	22
256 x 4	MCM93L422A	45	3-State	22
256 x 4	MCM93422	45	3-State	22
256 x 4	MCM93422A	35	3-State	22

# RAMs (continued)

## ECL 10K, 10KH RAMs

Organization	Part Number	Access Time (ns max)	No. of Output	Pins
8x2	MCM10143	15.3	O.E.	24
16x4	MC10H145	6	O.E.	16
16x4	MCM10145	15	O.E.	16
64x1	MCM10148	15	O.E.	16
128x1	MCM10147	15	O.E.	16
256x1	MCM10144	26	O.E.	16
256x1	MCM10152	15	O.E.	16
1024x1	MCM10146	29	O.E.	16
1024x1	MCM10415-20	20	O.E.	16
1024x1	MCM10415-15	15	O.E.	16
1024x1	MCM10415-10	10	O.E.	16
256x4	MCM10422-15	15	O.E.	24
256x4	MCM10422-10	10	O.E.	24
4096x1	MCM10470-25	25	O.E.	18
4096x1	MCM10470-15	15	O.E.	18
1024x4	MCM10474-25	25	O.E.	24
1024x4	MCM10474-15	15	O.E.	24
16384x1	MCM10480-20*	20	O.E.	20
4096x4	MCM10484-20*	20	O.E.	28

## ECL 100K RAMs

1024x1	MCM100415-10*	10	O.E.	16
256x4	MCM100422-10*	10	O.E.	24
4096x1	MCM100470-15*	15	O.E.	18
1024x4	MCM100474-15*	15	O.E.	24
16384x1	MCM100480-20*	20	O.E.	20
4096x4	MCM100484-20*	20	O.E.	28

# EPROMs

## MOS EPROMs

Organization	Part Number	Access Time (ns max)	Power Supplies	No. of Pins
8192x8	MCM68764C	450	+5 V	24
8192x8	MCM68766C	450	+5 V	24
8192x8	MCM68766C35	350	+5 V	24

Operating temperature ranges:

MOS - 0°C to 70°C

ECL - 0°C to 75°C

TTL - Military -55°C to +125°C, Commercial 0°C to 75°C

\* To be introduced.

(Not all speed selections shown)

# PROMs

## ECL PROMs

Organization	Part Number	Access Time (ns max)	No. of Pins
32x8	MCM10139	20	16
256x4	MCM10149	25	16
256x4	MCM10149A*	15	16

## TTL PROMs (3-State Outputs)

Organization	Part Number	Access Time (ns max)	No. of Pins
32x8	MCM27S19*	25	16
512x4	MCM7621	70	16
512x4	MCM7621A	60	16
512x8	MCM7641	70	24
512x8	MCM7641A	60	24
512x8	MCM7649	60	20
512x8	MCM7649A	45	20
512x8	MCM27S29*	35	20
512x8	MCM27S31*	35	24
512x8	MCM27S25*	See Note 3	24
512x8	MCM27S27*	See Note 3	22
1024x4	MCM7643	70	18
1024x4	MCM7643A	50	18
1024x8	MCM7681	70	24
1024x8	MCM7681A	50	24
1024x8	MCM27S181*	35	24 <sup>4</sup>
1024x8	MCM27S281*	35	24 <sup>2</sup>
1024x8	MCM27S35 <sup>5</sup> *	See Note 3	24
1024x8	MCM27S37 <sup>6</sup> *	See Note 3	24
2048x4	MCM7685	70	18
2048x4	MCM7685A	55	18
2048x8	MCM76161	70	24
2048x8	MCM76161A	50	24
2048x8	MCM27S191	35	24 <sup>4</sup>
2048x8	MCM27S291	35	24 <sup>2</sup>
2048x8	MCM27S45 <sup>5</sup>	See Note 3	24
2048x8	MCM27S47 <sup>6</sup>	See Note 3	24
4096x4	MCM76165	35	20

<sup>2</sup>300 mil package

<sup>3</sup>Registered Outputs - 20 ns max clock to output  
35 ns max address to clock setup time

<sup>4</sup>600 mil package

<sup>5</sup>Asynchronous register

<sup>6</sup>Synchronous register



**MOTOROLA Semiconductor Products Inc.**

# ROMs

## MOS STATIC ROMs (+5 Volts)

### Character Generators<sup>7</sup>

Organization	Part Number	Access Time (ns max)	No. of Pins
128 × (7 × 5)	MCM6670P	350	18
128 × (7 × 5)	MCM6674P	350	18
128 × (9 × 7)	MCM66700P	350	24
128 × (9 × 7)	MCM66710P	350	24
128 × (9 × 7)	MCM66714P	350	24
128 × (9 × 7)	MCM66720P	350	24
128 × (9 × 7)	MCM66730P	350	24
128 × (9 × 7)	MCM66734P	350	24
128 × (9 × 7)	MCM66740P	350	24
128 × (9 × 7)	MCM66750P	350	24
128 × (9 × 7)	MCM66760P	350	24
128 × (9 × 7)	MCM66770P	350	24
128 × (9 × 7)	MCM66780P	350	24
128 × (9 × 7)	MCM66790P	350	24

## MOS Binary ROMs (+5 Volts)

Organization	Part Number	Access Time (ns max)	No. of Pins
2048 × 8	MCM68A316EP	350	24
2048 × 8	MCM68A316EP91 <sup>8</sup>	350	24
4096 × 8	MCM68A332P	350	24
4096 × 8	MCM68A332P2 <sup>3</sup>	350	24
8192 × 8	MCM68364P35	350	24
8192 × 8	MCM68364P35-3 <sup>8</sup>	350	24
8192 × 8	MCM68364P25	250	24
8192 × 8	MCM68364P20	200	24
8192 × 8	MCM68365P25	250	24
8192 × 8	MCM68365P35	350	24
8192 × 8	MCM68366P25	250	24
8192 × 8	MCM68366P35	350	24
16384 × 8	MCM63128P15	150	28
16384 × 8	MCM63128P20	200	28
32768 × 8	MCM63256P15	150	28
32768 × 8	MCM63256P20	200	28

## CMOS ROMs (+5 Volts)

Organization	Part Number	Access Time (ns max)	No. of Pins
256 × 4	MCM14524	1200	16
2048 × 8	MCM65516P43	430	18
2048 × 8	MCM65516P43M <sup>8</sup>	430	18
2048 × 8	MCM65516P55	550	18

### Operating temperature ranges:

MOS — 0°C to 70°C

ECL — 0°C to 75°C

TTL — Military — 55°C to +125°C, Commercial 0°C to 75°C

\* To be introduced.

(Not all speed selections shown)

<sup>7</sup>Character generators include shifted and unshifted characters, ASCII alphanumeric control, math, Japanese British, German, European and French symbols.

<sup>8</sup>Standard Patterns for MOS ROMs:

MCM68A316EP91 — Universal Code Converter and Character Generator

MCM68A332P2 — Sine/Cosine Look-Up Table

MCM68364P35-3 — Log/Antilog Look-Up Table

MCM65516P43M — MC146805 Monitor Program



**MOTOROLA** Semiconductor Products Inc.



MOTOROLA

# SEMICONDUCTORS

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

## Advance Information

### 16,384-BIT DYNAMIC RAM

The MCM4517 is a 16,384-bit, high-speed, dynamic Random-Access Memory. Organized as 16,384 one-bit words and fabricated using HMOS high-performance, N-channel, silicon-gate technology. This new breed of 5-volt only dynamic RAM combines high performance with low cost and improved reliability.

By multiplexing row- and column-address inputs, the MCM4517 requires only seven address lines and permits packaging in standard 16-pin dual-in-line packages. Complete address decoding is done on chip with address latches incorporated. Data out is controlled by CAS allowing for greater system flexibility.

All inputs and outputs, including clocks, are fully TTL compatible. The MCM4517 incorporates a one-transistor cell design and dynamic storage techniques.

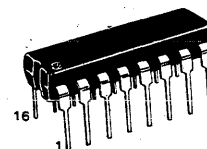
- Organized as 16,384 Words of 1 Bit
- Single +5 Volt Operation
- Fast 100 ns Operation
- Low Power Dissipation:
  - 170 mW Maximum (Active)
  - 14 mW Maximum (Standby)
- Maximum Access Time
  - MCM4517-10 — 100 ns
  - MCM4517-12 — 120 ns
  - MCM4517-15 — 150 ns
  - MCM4517-20 — 200 ns
- Three-State Data Output
- Internal Latches for Address and Data Input
- Early-Write Common I/O Output Capability
- 64K Compatible 128-cycle, 2 ms Refresh
- $\overline{RAS}$ -only Refresh Mode
- $\overline{CAS}$  Controlled Output
- Upward Pin Compatibility from the 16K RAM (MCM4116) to the 64K RAM (MCM6664)
- Allows Undershoot  $V_{IL\ min} = -2\ V$
- Hidden  $\overline{RAS}$  Only Refresh Capability

# MCM4517

## MOS

(N-CHANNEL, SILICON-GATE)

### 16,384-BIT DYNAMIC RAM

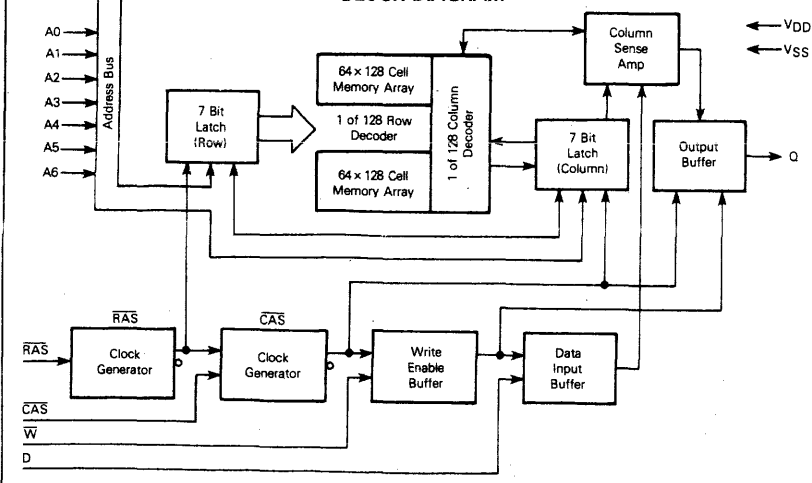


P SUFFIX  
PLASTIC PACKAGE  
CASE 648

#### PIN ASSIGNMENT

N/C	1	16	VSS
D	2	15	$\overline{CAS}$
$\overline{W}$	3	14	Q
$\overline{RAS}$	4	13	A6
A0	5	12	A3
A2	6	11	A4
A1	7	10	A5
VCC	8	9	N/C

#### BLOCK DIAGRAM



#### PIN NAMES

A0-A6	Address Input
D	Data In
Q	Data Out
$\overline{W}$	Read/Write Input
$\overline{RAS}$	Row Address Strobe
$\overline{CAS}$	Column Address Strobe
VCC	Power (+5 V)
VSS	Ground

This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid application of any voltage higher than maximum rated voltages to this high-impedance circuit.

This document contains information on a new product. Specifications and information herein are subject to change without notice.



MOTOROLA

SEMICONDUCTORS

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

Advance Information

64K BIT DYNAMIC RAM

The MCM6665B is a 65,536-bit, high-speed, dynamic Random-Access Memory. It is organized as 65,536 one-bit words and fabricated using HMOS high-performance N-channel silicon-gate technology, and is a yield enhanced version of our popular MCM6664A, featuring a smaller die size and redundancy.

By multiplexing row- and column-address inputs, the MCM6665B requires only eight address lines and permits packaging in standard 16-pin dual-in-line packages. Complete address decoding is done on chip with address latches incorporated. Data out is controlled by CAS allowing for greater system flexibility.

All inputs and outputs, including clocks, are fully TTL compatible. The MCM6665B incorporates a one-transistor cell design and dynamic storage techniques.

- Organized as 65,536 Words of 1 Bit
- Single 5 V Operation ( $\pm 10\%$ )
- Maximum Access Time
  - MCM6665B-15 = 150 ns
  - MCM6665-20 = 200 ns
- Low Power Dissipation
  - 302.5 mW Maximum (Active) MCM6665B-15
  - 22 nW Maximum (Standby)
- Three-State Data Output
- Internal Latches for Address and Data Input
- Early-Write Common I/O Capability
- 16K Compatible 128-Cycle, 2 ms Refresh
- RAS-only Refresh Mode
- CAS Controlled Output
- Upward Pin Compatible from the 16K RAM (MCM4116, MCM4517)
- Fast Page Mode Cycle Time

MCM6665B

MOS

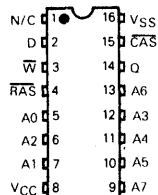
(N-CHANNEL, SILICON-GATE)

65,536-BIT DYNAMIC RANDOM ACCESS MEMORY



P SUFFIX PLASTIC PACKAGE CASE 648

PIN ASSIGNMENT

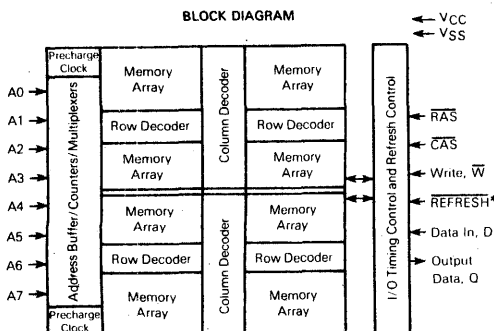


PIN NAMES

A0-A7	Address Input
D	Data In
Q	Data Out
W	Read/Write Input
RAS	Row Address Strobe
CAS	Column Address Strobe
VCC	Power (+5 V)
VSS	Ground

This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid application of any voltage higher than maximum rated voltages to this high-impedance circuit.

BLOCK DIAGRAM



\*Refresh Function Available on MCM6664B

This document contains information on a new product. Specifications and information herein are subject to change without notice.

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AD1-981-R1



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Product Preview

64K BIT DYNAMIC RAM

The MCM6665C is a 65,536-bit, high-speed, dynamic Random-Access Memory. Organized as 65,536 one-bit words and fabricated using HMOS high-performance N-channel silicon-gate technology, this new breed of 5-volt only dynamic RAM combines high performance with low cost and improved reliability.

By multiplexing row- and column-address inputs, the MCM6665C requires only eight address lines and permits packaging in standard 16-pin dual-in-line packages. Complete address decoding is done on chip with address latches incorporated. Data out is controlled by CAS, allowing for greater system flexibility.

All inputs and outputs, including clocks, are fully TTL compatible. The MCM6665C incorporates a one-transistor cell design and dynamic storage techniques.

- Organized as 65,536 Words of 1 Bit
- Single + V Operation ( $\pm 10\%$ )
- Full Power Supply Range Capabilities
- Maximum Access Time
  - MCM6665C-12 = 120 ns
  - MCM6665C-15 = 150 ns
  - MCM6665C-20 = 200 ns
- Low Power Dissipation
- Three-State Data Output
- Internal Latches for Address and Data Input
- Early-Write Common I/O Capability
- 16K Compatible 128-Cycle, 2 ms Refresh
- RAS-only Refresh Mode
- CAS Controlled Output
- Upward Pin Compatible from the 16K RAM (MCM4116, MCM4517)
- Fast Page Mode Cycle Time
- Redundancy Added

MCM6665C

MOS

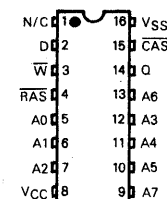
(N-CHANNEL, SILICON-GATE)

65, 536-BIT DYNAMIC RANDOM ACCESS MEMORY



P SUFFIX PLASTIC PACKAGE CASE 648

PIN ASSIGNMENT

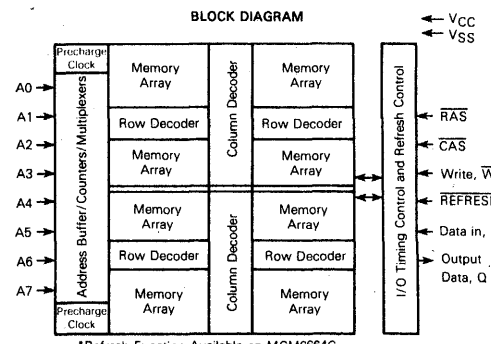


PIN NAMES

A0-A7	Address Input
D	Data In
Q	Data Out
W	Read/Write Input
RAS	Row Address Strobe
CAS	Column Address Strobe
VCC	Power (+5 V)
VSS	Ground

This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid application of any voltages to this high-impedance circuit.

BLOCK DIAGRAM



\*Refresh Function Available on MCM6664C

This document contains information on a new product. Specifications and herein are subject to change without notice.

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AD1-999



**MOTOROLA**

# SEMICONDUCTORS

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

## Product Preview

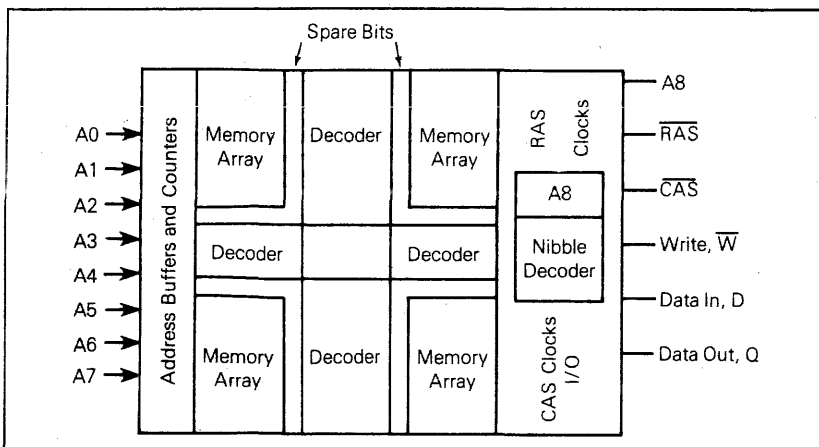
### 256K-BIT DYNAMIC RAM

The MCM6256 is a 262,144 bit, high-speed, dynamic Random Access Memory. Organized as 262,144 one-bit words and fabricated using Motorola's high-performance silicon-gate MOS (HMOS) technology, this new single +5 volt supply dynamic RAM combines high performance with low cost and improved reliability. The MCM6256 has the capability of using laser fuse redundancy and is manufactured using advanced direct-step on wafer photolithographic equipment.

By multiplexing row and column address inputs, the MCM6256 requires only nine address lines and permits packaging in standard 16-pin 300 mil wide dual-in-line packages. Complete address decoding is done on-chip with address latches incorporated. Data out (Q) is controlled by  $\overline{\text{CAS}}$  allowing greater system flexibility.

All inputs and outputs, including clocks, are fully TTL compatible. The MCM6256 incorporates a one transistor cell design and dynamic storage techniques. In addition to the  $\overline{\text{RAS}}$ -only refresh mode, a  $\overline{\text{CAS}}$  before  $\overline{\text{RAS}}$  automatic refresh is available. Another special feature of the MCM6256 is nibble mode, allowing the user to serially access 4 bits of data at a high data rate. Nibble mode address is controlled by the addresses on pin 1 (A8 row and A8 column).

- Organized as 262,144 Words of 1 Bit
- Single +5 Volt Operation ( $\pm 10\%$ )
- Maximum Access Time:
  - MCM6256-10 = 100 ns
  - MCM6256-12 = 120 ns
  - MCM6256-15 = 150 ns
- Low Power Dissipation:
  - 70 mA maximum (Active) MCM6256-10
  - 4 mA maximum (Standby)
- Three-State Data Output
- Early-Write Common I/O Capability
- 256 Cycle, 4 ms Refresh
- $\overline{\text{RAS}}$ -Only Refresh Mode
- Automatic ( $\overline{\text{CAS}}$  before  $\overline{\text{RAS}}$ ) Refresh Mode
- Fast Nibble Mode on Read and Write Cycles
  - 20 ns Access Time
  - 40 ns Cycle Time

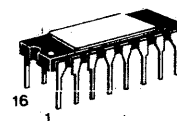


# MCM6256

## MOS

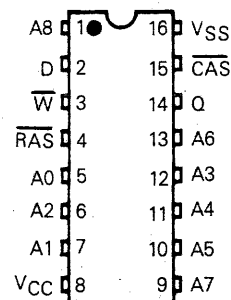
(N-CHANNEL, SILICON-GATE)

### 262,144 BIT DYNAMIC RANDOM ACCESS MEMORY



L SUFFIX  
CERAMIC PACKAGE  
CASE 690

#### PIN ASSIGNMENT



#### PIN NAMES

A0-A8	Address Input
D	Data In
Q	Data Out
$\overline{\text{W}}$	Read/Write Input
$\overline{\text{RAS}}$	Row Address Strobe
$\overline{\text{CAS}}$	Column Address Strobe
VCC	Power (+5 V)
VSS	Ground

This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid application of any voltage higher than maximum rated voltages to this high-impedance circuit.



**MOTOROLA**

# SEMICONDUCTORS

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

## Advance Information

### FAST 16K BIT STATIC RAM

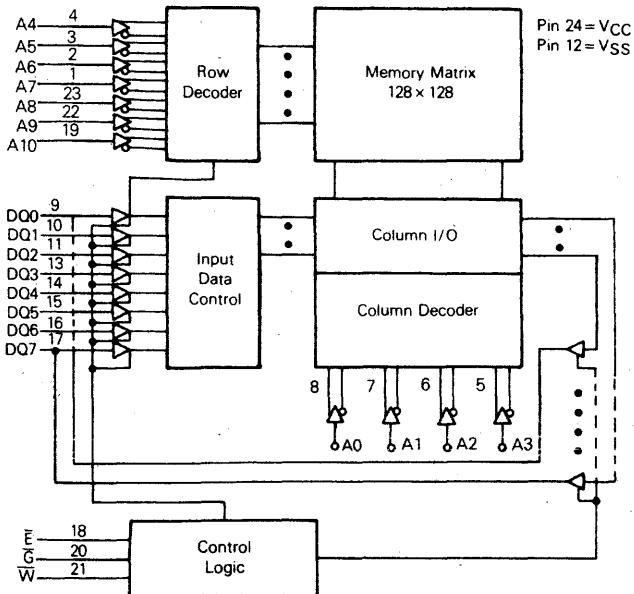
The MCM2016H is a 16,384-bit Static Random Access Memory organized as 2048 words by 8 bits, fabricated using Motorola's High-performance silicon-gate MOS (HMOS) technology. It uses an innovative design approach which combines the ease-of-use features of fully static operation (no external clocks or timing strobes required) with the reduced standby power dissipation associated with clocked memories. To the user this means low standby power dissipation without the need for address setup and hold times, nor reduced data rates due to cycle times that are no longer than access times. Perfect for cache and sub-100 ns buffer memory systems, this high speed static RAM is intended for applications that demand superior performance and reliability.

Chip Enable ( $\bar{E}$ ) controls the power-down feature. It is not a clock but rather a chip control that affects power consumption. In less than a cycle time after chip enable ( $\bar{E}$ ) goes high, the part automatically reduces its power requirements and remains in this low-power standby mode as long as the chip enable ( $\bar{E}$ ) remains high. This feature provides significant system-level power savings.

The MCM2016H is in a 24-pin dual-in-line 600 mil wide package with the industry standard JEDEC approved pinout and is pinout compatible with the industry standard 16K EPROM/ROM. A 24 pin dual-in-line 300 mil wide package will also be available.

- Single +5 Volt Operation ( $\pm 10\%$ )
- Fully Static: No Clock or Timing Strobe Required
- Fast Access Time: MCM2016H-45 - 45 ns (max)  
MCM2016H-55 - 55 ns (max)  
MCM2016H-70 - 70 ns (max)
- Power Dissipation: 120 mA Maximum (Active)  
20 mA Maximum (Standby)
- Three-State Output

### BLOCK DIAGRAM

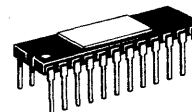


# MCM2016H

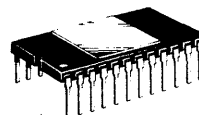
## MOS

(N-CHANNEL, SILICON-GATE)

### 2,048 x 8 BIT STATIC RANDOM ACCESS MEMORY



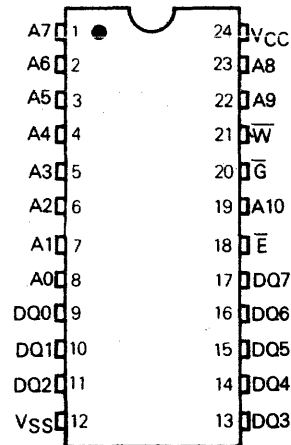
**Y SUFFIX**  
CERAMIC PACKAGE  
CASE 716-08



**L SUFFIX**  
CERAMIC PACKAGE  
CASE 716-06

**P SUFFIX PLASTIC PACKAGE**  
ALSO AVAILABLE - CASE 709

### PIN ASSIGNMENTS



### PIN NAMES

A0-A10	.....	Address Input
DQ0-DQ7	.....	Data Input/Output
W	.....	Write Enable
G	.....	Output Enable
E	.....	Chip Enable
VCC	.....	Power (+5 V)
VSS	.....	Ground

Motorola Semiconductor

MEMORY

This document contains information on a new product. Specifications and information herein are subject to change without notice.

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ADI-951-R1





**MOTOROLA**

# SEMICONDUCTORS

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

## Advance Information

### FAST 16K BIT STATIC RAM

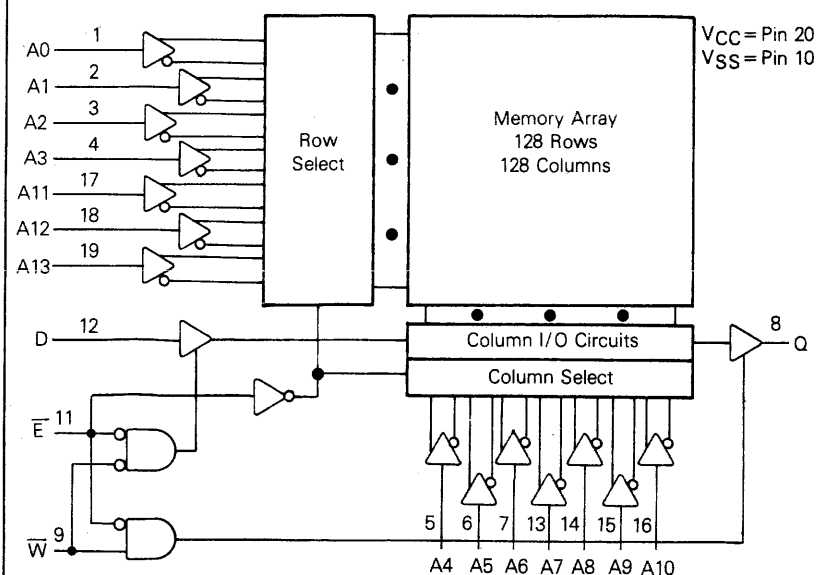
The MCM2167H is a 16,384-bit Static Random Access Memory organized as 16,384 words by 1 bit, fabricated using Motorola's High-performance silicon-gate MOS (HMOS) technology. It uses an innovative design approach which combines the ease-of-use features of fully static operation (no external clocks or timing strobes required) with the reduced standby power dissipation associated with clocked memories. To the user this means low standby power dissipation without the need for address setup and hold times, nor reduced data rates due to cycle times that are no longer than access times. Perfect for cache and sub-100 ns buffer memory systems, this high speed static RAM is intended for applications demanding superior performance and reliability.

Chip Enable ( $\bar{E}$ ) controls the power-down feature. It is not a clock but rather a chip control that affects power consumption. In less than a cycle time after Chip Enable ( $\bar{E}$ ) goes high, the part automatically reduces its power requirements and remains in this low-power standby mode as long as the Chip Enable ( $\bar{E}$ ) remains high. This feature provides significant system-level power savings.

The MCM2167H is in a 20 pin dual-in-line package with the industry standard pinout. It is TTL compatible in all respects. The data out has the same polarity as the input data. A data input and a separate three-state output provide flexibility and allow easy OR-ties.

- Single +5 V Operation ( $\pm 10\%$ )
- Fully Static Memory — No Clock or Timing Strobe Required
- Fast Access Time: MCM2167H-35 — 35 ns Max.  
MCM2167H-45 — 45 ns Max.
- Power Dissipation: 120 mA Maximum (Active)  
20 mA Maximum (Standby)
- Low Power Also Available: MCM21L67H
- Automatic Power Down

### BLOCK DIAGRAM

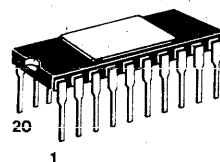


# MCM2167H

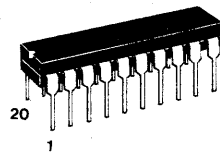
## MOS

(N-CHANNEL, SILICON-GATE)

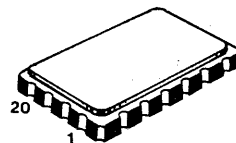
### 16,384-BIT STATIC RANDOM ACCESS MEMORY



**L SUFFIX**  
CERAMIC PACKAGE  
CASE 729

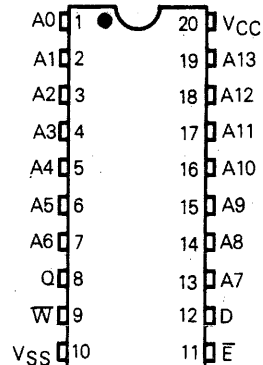


**P SUFFIX**  
PLASTIC PACKAGE  
CASE 738



**Z SUFFIX**  
LEADLESS CHIP CARRIER  
CASE 756B

### PIN ASSIGNMENT



### PIN NAMES

A0-A13	Address Input
W	Write Enable
$\bar{E}$	Chip Enable
D	Data Input
Q	Data Output
V <sub>CC</sub>	Power (+5 V)
V <sub>SS</sub>	Ground

This document contains information on a new product. Specifications and information herein are subject to change without notice.

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ADI-980

4053

Motorola Semiconductor

MEMORY



MOTOROLA

# SEMICONDUCTORS

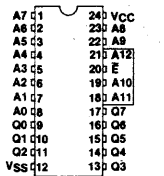
3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

## Nonvolatile Memories

### 24 PIN-COMPATIBLE 64K ROM FAMILY

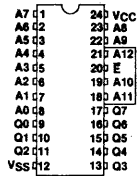
### 24 PIN-COMPATIBLE 64K ROM FAMILY (CONTINUED)

#### MCM68364



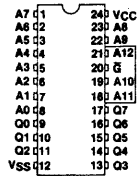
- EDGE ACTIVATED
- POWER DOWN
- ADDRESS LATCH

#### MCM68365



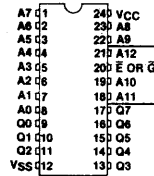
- FULLY STATIC
- POWER DOWN
- tACC = tCYCLE

#### MCM68366

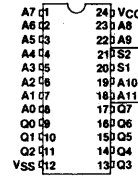


- FULLY STATIC
- OUTPUT ENABLE
- FAST ACCESS

#### MCM68364/5/6

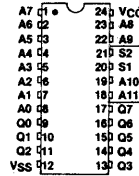


#### MCM68367



- FULLY STATIC
- PAGE MODE

#### MCM68368



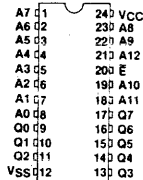
- FULLY STATIC
- BANK SELECT

#### MCM68365

##### CHARACTERISTICS

- 8192 x 8 MASK PROGRAMMABLE ROM
- SINGLE +5 VOLT 10% SUPPLY
- PIN COMPATIBLE WITH 8K, 16K, AND 32K ROMs
- FULLY STATIC OPERATION
- AUTOMATIC POWER DOWN
- LOW POWER DISSIPATION  
225 mW ACTIVE (TYPICAL)  
30 mW STANDBY (TYPICAL)
- MAXIMUM ACCESS TIMES:  
68365C30 — 300 ns  
68365C35 — 350 ns

##### PINOUT

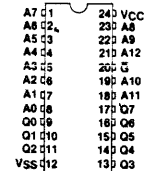


#### MCM68366

##### CHARACTERISTICS

- 8192 x 8 MASK PROGRAMMABLE ROM
- SINGLE +5 VOLT 10% SUPPLY
- INDUSTRY STANDARD 24 PIN PACKAGE
- FULLY STATIC OPERATION (NO CLOCKS OR STROBES)
- FAST 150 ns ACCESS TIME FROM OUTPUT ENABLE
- LOW POWER DISSIPATION  
225 mW ACTIVE (TYPICAL)
- FULLY TTL COMPATIBLE
- MASK PROGRAMMABLE OUTPUT ENABLE
- PIN COMPATIBLE WITH MCM68766 (64K EPROM)

##### PINOUT

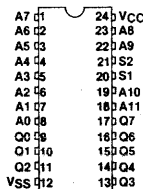


#### MCM68367

##### CHARACTERISTICS

- 8K x 8 HMOS MASK PROGRAMMABLE ROM
- PAGE MODE — ACCESS TOP HALF OR BOTTOM HALF OF THE 4K BYTES
- APPEARS TO OUTSIDE WORLD AS 4K x 8 STANDARD ROM
- FULLY STATIC OPERATION (NO CLOCKS)
- SUPPLY CURRENT  
100 mA ACTIVE (MAX)
- MAXIMUM ACCESS TIME — 450 ns
- PROGRAM LAYER LATE IN PROCESS

##### PINOUT

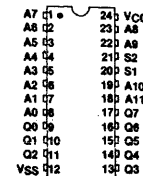


#### MCM68368

##### CHARACTERISTICS

- 8K x 8 HMOS MASK PROGRAMMABLE ROM
- MEMORY IS DIVIDED INTO EIGHT 1K x 8 BYTES WHICH ARE SELECTED BY INDIRECT ADDRESSING
- APPEARS TO OUTSIDE WORLD AS 4K x 8 ROM STANDARD PACKAGE
- FULLY STATIC OPERATION (NO CLOCKS)
- SUPPLY CURRENT  
125 mA ACTIVE (MAX)
- MAX ACCESS TIME — 450 ns
- PROGRAM LAYER LATE IN PROCESS

##### PINOUT



## 28 PIN-COMPATIBLE ROM FAMILY

### MCM68369

N/C	1	28	VCC
A12	2	27	S1
A7	3	26	S2
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	S3
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

● 64K  
(8K × 8)

### MCM68370

N/C	1	28	VCC
A12	2	27	S1
A7	3	26	S2
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	E
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

● 64K  
(8K × 8)

### MCM63128

N/C	1	28	VCC
A12	2	27	S
A7	3	26	A13
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	E
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

● 128K  
(16K × 8)

### MCM63256

N/C	1	28	VCC
A12	2	27	A14
A7	3	26	A13
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	E
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

● 256K  
(32K × 8)

### MCM68369

#### CHARACTERISTICS

- 8K × 8 NMOS MASK PROGRAMMABLE ROM
- INDUSTRY STANDARD 28 PIN PACKAGE
- FULLY STATIC OPERATION (NO CLOCKS OR STROBES)
- SUPPLY CURRENT
  - 100 mA ACTIVE (MAXIMUM)
  - 15 mA STANDBY (MAXIMUM)
- MAXIMUM ACCESS TIME — 350 ns
- MASK PROGRAMMABLE CHIP SELECTS AND OUTPUT ENABLE
- PROGRAM LAYER LATE IN PROCESS

#### PINOUT

NC	1	28	VCC
A12	2	27	S1
A7	3	26	S2
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	S3
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

### MCM68370

#### CHARACTERISTICS

- 8K × 8 NMOS MASK PROGRAMMABLE ROM
- INDUSTRY STANDARD 28 PIN PACKAGE
- FULLY STATIC OPERATION (NO CLOCKS OR STROBES)
- SUPPLY CURRENT
  - 100 mA ACTIVE (MAXIMUM)
  - 15 mA STANDBY (MAXIMUM)
- MAXIMUM ACCESS TIME — 350 ns
- MASK PROGRAMMABLE CHIP ENABLE, OUTPUT ENABLE, AND CHIP SELECTS
- PROGRAM LAYER LATE IN PROCESS

#### PINOUT

NC	1	28	VCC
A12	2	27	S1
A7	3	26	S2
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	E
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

### MCM63128

#### CHARACTERISTICS

- 16K × 8 HMOS MASK PROGRAMMABLE ROM
- INDUSTRY STANDARD 28 PIN PACKAGE
- FULLY STATIC OPERATION (NO CLOCKS OR STROBES)
- SUPPLY CURRENT
  - 100 mA ACTIVE (MAXIMUM)
  - 15 mA STANDBY (MAXIMUM)
- MAXIMUM ACCESS TIME — 200 ns
- MASK PROGRAMMABLE CHIP AND OUTPUT ENABLES
- PROGRAM LAYER LATE IN PROCESS
- THE FUNCTIONS E AND S ON PINS 20 AND 27 CAN BE INTERCHANGED AT USER OPTION
- G CAN BE ORed OR ANDed WITH S OR E AT USER OPTION

#### PINOUT

NC	1	28	VCC
A12	2	27	S
A7	3	26	A13
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	E
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3

### MCM63256

#### CHARACTERISTICS

- 32,768x8 HMOS MASK PROGRAMMABLE ROM
- INDUSTRY STANDARD 28 PIN PACKAGE
- FULLY STATIC OPERATION (NO CLOCKS OR STROBES)
- SUPPLY CURRENT
  - 100 mA ACTIVE (MAXIMUM)
  - 15 mA STANDBY (MAXIMUM)
- MAXIMUM ACCESS TIME — 150 ns
- MASK PROGRAMMABLE CHIP AND OUTPUT ENABLES
- ADDRESS (A14) IS USER SELECTABLE FOR EITHER PIN 27 OR PIN 1
- PROGRAM LAYER LATE IN PROCESS

#### PINOUT

NC	1	28	VCC
A12	2	27	A14
A7	3	26	A13
A6	4	25	A8
A5	5	24	A9
A4	6	23	A11
A3	7	22	G
A2	8	21	A10
A1	9	20	E
A0	10	19	Q7
Q0	11	18	Q6
Q1	12	17	Q5
Q2	13	16	Q4
VSS	14	15	Q3



**MOTOROLA** Semiconductor Products Inc.

**MOTOROLA SEMICONDUCTORS**

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

**MCM63128**

**Advance Information**

**128K BIT READ ONLY MEMORY**

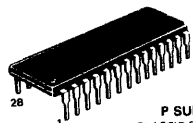
The MCM63128 is a MOS mask programmable, byte-oriented, Read-Only Memory (ROM). The MCM63128 is organized as 16x8 and is fabricated using Motorola's high performance N-channel silicon gate technology (HMOS). This device is designed to provide maximum circuit density and reliability with the highest possible performance. It remains fully compatible with TTL inputs and outputs while maintaining low power dissipation and wide operating margins. The MCM63128 also contains circuitry for current surge suppression which maintains the chip in an internal deselect mode until V<sub>CC</sub> approaches 2.5 volts, at which time the chip is internally selected.

There are numerous logical nor or NAND combinations between Chip Enable (E), Chip Select (S), and Output Enable (G) that tri state the device. This feature is selected by the user and placed into effect with the mask programming. The active level of the Chip Enable (E), Chip Select (S), and the Output Enable (G), along with the memory contents, are defined by the user.

The Chip Enable input controls the automatic power down feature which deselects the outputs and reduces the power consumption.

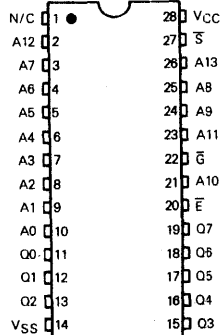
- Single + 5 Volt (± 10%) Supply
- Fully Static Periphery - No Clocking Required on Chip Enable
- Power Dissipation  
100 mA Active (Maximum) (Unloaded)  
15 mA Standby (Maximum)
- Program Layer Late in Process for Quick Turnaround Time
- 150 ns - MCM63128P15  
200 ns - MCM63128P20
- Maximum Access from Address and Chip Enable  
60 ns - MCM63128P15  
80 ns - MCM63128P20
- Active Level for Chip Enable and Output Enable are User Selectable
- 28-Pin JEDEC Standard Package and Pinout

**HMOS (N-CHANNEL, SILICON GATE) 16,384 x 8 BIT READ ONLY MEMORY**



P SUFFIX PLASTIC PACKAGE CASE 710

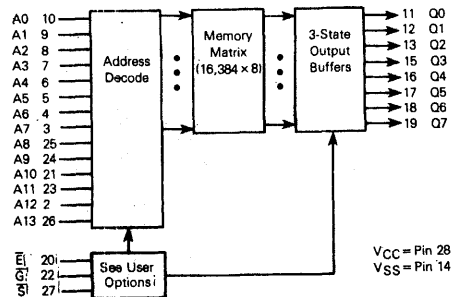
**PIN ASSIGNMENT**



**PIN NAMES**

A0-A13	Address
E	Chip Enable
G	Output Enable
S	Chip Select
Q0-Q7	Data Output
V <sub>CC</sub>	+5 V Power Supply
V <sub>SS</sub>	Ground

**BLOCK DIAGRAM**



V<sub>CC</sub> = Pin 28  
V<sub>SS</sub> = Pin 14

This document contains information on a new product. Specifications and information herein are subject to change without notice.

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**MOTOROLA SEMICONDUCTORS**

3501 ED BLUESTEIN BLVD., AUSTIN, TEXAS 78721

**MCM63256**

**Advance Information**

**256K BIT READ ONLY MEMORY**

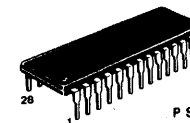
The MCM63256 is a MOS mask programmable byte-oriented, Read-Only Memory (ROM). The MCM63256 is organized as 32K x 8 and is fabricated using Motorola's high performance N-channel silicon gate technology (HMOS). This device is designed to provide maximum circuit density and reliability with the highest possible performance. It remains fully compatible with TTL inputs and outputs while maintaining low power dissipation and wide operating margins. The MCM63256 also contains circuitry for current surge suppression which maintains the chip in an internal deselect mode until V<sub>CC</sub> approaches 2.5 volts, at which time the chip is internally selected.

The active level of the Chip Enable and the Output Enable, along with the memory contents, are defined by the user. The user can also define the pinout assignment for address (A14) to either pin 27 or pin 1.

The Chip Enable input controls the automatic power down feature which deselects the outputs and reduces the power consumption.

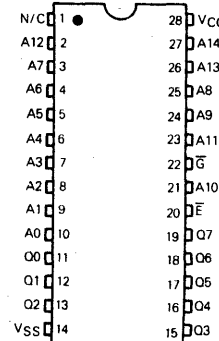
- Single + 5 Volt (± 10%) Supply
- Fully Static Periphery - No Clocking Required on Chip Enable
- Power Dissipation  
100 mA Active (Maximum) (Unloaded)  
15 mA Standby (Maximum)
- Program Layer Late in Process for Quick Turnaround Time
- Maximum Access from Address and Chip Enable  
150 ns - MCM63256P15  
200 ns - MCM63256P20
- Maximum Access from Output Enable  
60 ns - MCM63256P15  
80 ns - MCM63256P20
- Address (A14) is User Selectable for Either Pin 27 or Pin 1
- Active Level for Chip Enable and Output Enable are User Selectable
- 28-Pin JEDEC Standard Package and Pinout

**HMOS (N-CHANNEL, SILICON GATE) 32,768 x 8 BIT READ ONLY MEMORY**



P SUFFIX PLASTIC PACKAGE CASE 710

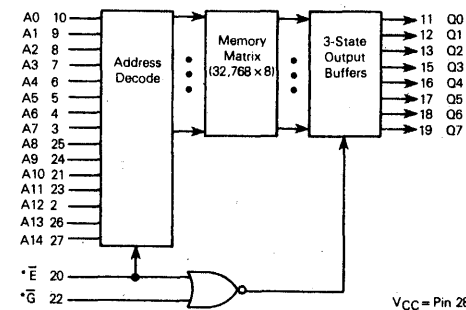
**PIN ASSIGNMENT**



**PIN NAMES**

A0-A14	Address
E	Chip Enable
G	Output Enable
Q0-Q7	Data Output
V <sub>CC</sub>	+5 V Power Supply
V <sub>SS</sub>	Ground

**BLOCK DIAGRAM**



V<sub>CC</sub> = Pin 28  
V<sub>SS</sub> = Pin 14

\*Active level defined by the user.

This document contains information on a new product. Specifications and information herein are subject to change without notice.

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ADI-989

## Introduction

### GENERAL

This generic Schottky PROM family by National Semiconductor makes available to the industry one of the widest selections in sizes and organizations. Four-bit wide PROMs are provided with 256 to 4096 words in pin compatible 16 and 18 pin dual-in-line packages. The 8-bit wide devices range from 32 to 4096 words in a variety of packages. Being 'generic,' all PROMs share a common programming algorithm.

### TITANIUM-TUNGSTEN FUSES

National's new Programmable Read-Only Memories (PROMs) feature titanium-tungsten (Ti:W) fuse links designed to program efficiently with only 10.5 Volts applied. The high performance and reliability of these PROMs are the result of fabrication by a Schottky bipolar process, of which the titanium-tungsten metalization is an integral part, and the use of an on-chip programming circuit.

A major advantage of the titanium-tungsten fuse technology is the low programming voltage of the fuse links. At 10.5 Volts, this virtually eliminates the need for guard-ring devices and wide spacings required for other fuse technologies. Care is taken, however, to minimize voltage drops across the die and to reduce parasitics. The device is designed to ensure that worst-case fuse operating current is low enough for reliable long-term operation. The Darlington programming circuit is liberally designed to insure adequate power density for blowing the fuse links. The complete circuit design is optimized to provide high performance over the entire operating ranges of  $V_{CC}$  and temperature.

### TESTABILITY

The Schottky PROM die includes extra rows and columns of fusable links for testing the programmability of each chip. These test fuses are placed at the worst-case chip locations to provide the highest possible confidence in the programming tests in the final product. A ROM pattern is also permanently fixed in the additional circuitry and coded to provide a parity check of input address levels. These and other test circuits are used to test for correct operation of the row and column-select circuits and functionality of input and enable gates. All test circuits are available at both wafer and assembled device levels to allow 100% functional and parametric testing at every stage of the test flow.

### RELIABILITY

As with all National products, the Ti:W PROMs are subjected to an on-going reliability evaluation by the Reliability Assurance Department. These evaluations employ accelerated life tests, including dynamic high-temperature operating life, temperature-humidity life, temperature cycling, and thermal shock. To date, nearly 7.4 million Schottky Ti:W PROM device hours have been logged, with samples in Epoxy B molded DIP (N-package) and CERDIP (J-package). Device performance in all package configurations is excellent.



## Bipolar PROM Selection Guide

Size (Bits)	Organization	DIP (Pins)	Part Number	TAA (Max) in nS	TEA (Max) in nS	ICC (Max) in mA	Temperature Celsius	
256	32 x 8	OC	16 DM54S188	45	30	110	-55 to +125	
	32 x 8	OC	16 DM74S188	35	20	110	0 to +70	
	32 x 8	TS	16 DM54S288	45	30	110	-55 to +125	
	32 x 8	TS	16 DM74S288	35	20	110	0 to +70	
1024	256 x 4	OC	16 DM54S387	60	30	130	-55 to +125	
	256 x 4	OC	16 DM74S387	50	25	130	0 to +70	
	256 x 4	TS	16 DM54S287	60	30	130	-55 to +125	
	256 x 4	TS	16 DM74S287	50	25	130	0 to +70	
2048	512 x 4	OC	16 DM54S570	65	35	130	-55 to +125	
	512 x 4	OC	16 DM74S570	55	30	130	0 to +70	
	512 x 4	TS	16 DM54S571	65	35	130	-55 to +125	
	512 x 4	TS	16 DM74S571	55	30	130	0 to +70	
	512 x 4	OC	16 DM54S570A	60	35	130	-55 to +125	
	512 x 4	OC	16 DM74S570A	45	25	130	0 to +70	
	512 x 4	TS	16 DM54S571A	60	35	130	-55 to +125	
	512 x 4	TS	16 DM74S571A	45	25	130	0 to +70	
	512 x 4	TS	16 DM54S571B	50	35	130	-55 to +125	
	512 x 4	TS	16 DM74S571B	35	25	130	0 to +70	
	256 x 8	TS	20 DM54LS471	70	35	100	-55 to +125	
	256 x 8	TS	20 DM74LS471	60	30	100	0 to +70	
	4096	512 x 8	OC	20 DM54S473	75	35	155	-55 to +125
512 x 8		OC	20 DM74S473	60	30	155	0 to +70	
512 x 8		TS	20 DM54S472	75	35	155	-55 to +125	
512 x 8		TS	20 DM74S472	60	30	155	0 to +70	
512 x 8		OC	20 DM54S473A	60	35	155	-55 to +125	
512 x 8		OC	20 DM74S473A	45	25	155	0 to +70	
512 x 8		TS	20 DM54S472A	60	35	155	-55 to +125	
512 x 8		TS	20 DM74S472A	45	25	155	0 to +70	
512 x 8		TS	20 DM54S472B	50	35	155	-55 to +125	
512 x 8		TS	20 DM74S472B	35	25	155	0 to +70	
512 x 8		OC	24 DM54S475	75	40	170	-55 to +125	
512 x 8		OC	24 DM74S475	65	35	170	0 to +70	
512 x 8		TS	24 DM54S474	75	40	170	-55 to +125	
512 x 8		TS	24 DM74S474	65	35	170	0 to +70	
512 x 8		OC	24 DM54S475A	60	35	170	-55 to +125	
512 x 8		OC	24 DM74S475A	45	25	170	0 to +70	
512 x 8		TS	24 DM54S474A	60	35	170	-55 to +125	
512 x 8		TS	24 DM74S474A	45	25	170	0 to +70	
512 x 8		TS	24 DM54S474B	50	35	170	-55 to +125	
512 x 8		TS	24 DM74S474B	35	25	170	0 to +70	
512 x 8		REG	24*	DM77SR474	40**	30	185	-55 to +125
512 x 8		REG	24*	DM87SR474	35**	25	185	0 to +70
512 x 8		REG	24*	DM77SR476	40**	30	185	-55 to +125
512 x 8		REG	24*	DM77SR25	40**	30	185	-55 to +125
512 x 8		REG	24*	DM87SR476	35**	25	185	0 to +70
512 x 8		REG	24*	DM87SR25	35**	25	185	0 to +70
1024 x 4		OC	18	DM54S572	75	45	140	-55 to +125
1024 x 4		OC	18	DM74S572	60	35	140	0 to +70
1024 x 4		TS	18	DM54S573	75	45	140	-55 to +125
1024 x 4		TS	18	DM74S573	60	35	140	0 to +70
1024 x 4	OC	18	DM54S572A	60	35	140	-55 to +125	
1024 x 4	OC	18	DM74S572A	45	25	140	0 to +70	
1024 x 4	TS	18	DM54S573A	60	35	140	-55 to +125	
1024 x 4	TS	18	DM74S573A	45	25	140	0 to +70	
1024 x 4	TS	18	DM54S573B	50	35	140	-55 to +125	
1024 x 4	TS	18	DM74S573B	35	25	140	0 to +70	

\* - 24 Pin Narrow Dual Inline Package

\*\* - Set up Time

 National Semiconductor

# Bipolar PROM Selection Guide



Size (Bits)	Organization	DIP (Pins)	Part Number	TAA (Max) in nS	TEA (Max) in nS	ICC (Max) in mA	Temperature Celsius		
8192	1024 x 8	OC	24	DM77S180	75	35	170	-55 to +125	
	1024 x 8	TS	24*	DM77S280	75	35	170	-55 to +125	
	1024 x 8	OC	24	DM87S180	55	30	170	0 to +70	
	1024 x 8	TS	24*	DM87S280	55	30	170	0 to +70	
	1024 x 8	OC	24	DM77S181	75	35	170	-55 to +125	
	1024 x 8	TS	24*	DM77S281	75	35	170	-55 to +70	
	1024 x 8	OC	24	DM87S181	55	30	170	0 to +70	
	1024 x 8	TS	24*	DM87S281	55	30	170	0 to +70	
	1024 x 8	TS	24	DM77LS181	175	70	100	-55 to +125	
	1024 x 8	TS	24	DM87LS181	120	50	100	0 to +70	
	1024 x 8	TS	24	DM77S181A	65	35	170	-55 to +125	
	1024 x 8	TS	24	DM87S181A	45	30	170	0 to +70	
	1024 x 8	REG.	24*	DM77SR181	50**	30	175	-55 to +125	
	1024 x 8	REG.	24*	DM87SR181	40**	25	175	0 to +70	
	2048 x 4	OC	18	DM77S184	75	35	140	-55 to +125	
	2048 x 4	OC	18	DM87S184	55	30	140	0 to +70	
	2048 x 4	TS	18	DM77S185	75	35	140	-55 to +125	
	2048 x 4	TS	18	DM87S185	55	30	140	0 to +70	
	2048 x 4	TS	18	DM77S185A	60	30	140	-55 to +125	
	2048 x 4	TS	18	DM87S185A	45	25	140	0 to +70	
	2048 x 4	TS	18	DM77S185B	50	30	140	-55 to +125	
	2048 x 4	TS	18	DM87S185B	35	25	140	0 to +70	
	16384	2048 x 8	OC	24	DM77S190	80	40	175	-55 to +125
		2048 x 8	TS	24*	DM77S290	80	40	175	-55 to +125
2048 x 8		OC	24	DM87S190	65	30	175	0 to +70	
2048 x 8		TS	24*	DM87S290	65	30	175	0 to +70	
2048 x 8		OC	24	DM77S191	80	40	175	-55 to +125	
2048 x 8		TS	24*	DM77S291	80	40	175	-55 to +125	
2048 x 8		OC	24	DM87S191	65	30	175	0 to +70	
2048 x 8		TS	24*	DM87S291	65	30	175	0 to +70	
2048 x 8		TS	24	DM77S191A	60	35	175	-55 to +125	
2048 x 8		TS	24	DM87S191A	45	30	175	0 to +70	
2048 x 8		TS	24	DM77S191B	50	30	175	-55 to +125	
2048 x 8		TS	24	DM87S191B	35	25	175	0 to +70	
4096 x 4		TS	20	DM77S195A	60	30	170	-55 to +125	
4096 x 4		TS	20	DM87S195A	45	25	170	0 to +70	
4096 x 4		TS	20	DM77S195B	50	30	170	-55 to +125	
4096 x 4		TS	20	DM87S195B	35	25	170	0 to +70	
32768	4096 x 8	TS	24	DM77S321	65	35	185	-55 to +125	
	4096 x 8	TS	24	DM87S321	55	30	185	0 to +70	
	4096 x 8	TS	24*	DM77S421	65	35	185	-55 to +125	
	4096 x 8	TS	24*	DM87S421	55	30	185	0 to +70	

### Absolute Maximum Ratings (Note 1)

Supply Voltage (Note 2)	-0.5 to +7.0V
Input Voltage (Note 2)	-1.2 to +5.5V
Output Voltage (Note 2)	-0.5 to +5.5V
Storage Temperature	-65 to +150C
Lead Temperature (10 seconds)	300C

- \* - 24 Pin Narrow Dual Inline Package
- \*\* - Set up Time

### Operating Conditions

	Min	Max	Units
Supply Voltage (V <sub>CC</sub> )			
Military	4.50	5.50	V
Commercial	4.75	5.25	V
Ambient Temperature (T <sub>A</sub> )			
Military	-55	+125	°C
Commercial	0	+70	°C
Logical "0" Input Voltage	0	0.8	V
Logical "1" Input Voltage	2.0	5.5	V

**Note 1:** Absolute maximum ratings are those values beyond which the device may be permanently damaged. They do not mean that the device may be operated at these values.

**Note 2:** These limits do not apply during programming. For the programming ratings, refer to the programming instructions.

## DC Electrical Characteristics (Note 3)

**National Semiconductor**

DM77/87S321, DM77/87S421, DM77/87S321A, DM77/87S421A

Symbol	Parameter	Conditions	DM77S321/421			DM87S321/421			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
$I_{IL}$	Input Load Current	$V_{CC} = \text{Max.}, V_{IN} = 0.45\text{V}$		-80	-250		-80	-250	$\mu\text{A}$
$I_{IH}$	Input Leakage Current	$V_{CC} = \text{Max.}, V_{IN} = 2.7\text{V}$			25			25	$\mu\text{A}$
		$V_{CC} = \text{Max.}, V_{IN} = 5.5\text{V}$			1.0			1.0	$\text{mA}$
$V_{OL}$	Low Level Output Voltage	$V_{CC} = \text{Min.}, I_{OL} = 12\text{mA}$		0.35	0.50		0.35	0.45	V
$V_{IL}$	Low Level Input Voltage				0.80			0.80	V
$V_{IH}$	High Level Input Voltage		2.0			2.0			V
$V_C$	Input Clamp Voltage	$V_{CC} = \text{Min.}, I_{IN} = -18\text{mA}$		-0.8	-1.2		-0.8	-1.2	V
$C_I$	Input Capacitance	$V_{CC} = 5.0, V_{IN} = 2.0\text{V}$ $T_A = 25^\circ\text{C}, 1\text{MHz}$		4.0			4.0		$\text{pF}$
$C_O$	Output Capacitance	$V_{CC} = 5.0\text{V}, V_O = 2.0\text{V}$ $T_A = 25^\circ\text{C}, 1\text{MHz}, \text{Outputs Off}$		6.0			6.0		$\text{pF}$
$I_{CC}$	Power Supply Current	$V_{CC} = \text{Max.},$ All Outputs Open		135	185		135	185	$\text{mA}$

### TRI-STATE Parameters

$I_{OS}$	Short Circuit Output Current	$V_O = 0\text{V}, V_{CC} = \text{Max.}$ (Note 4)	-20		-70	-20		-70	$\text{mA}$
$I_{OZ}$	Output Leakage (TRI-STATE)	$V_{CC} = \text{Max.}, V_O = 0.45 \text{ to } 2.4\text{V}$ Chip Disabled	-50		+50	-50		+50	$\mu\text{A}$
$V_{OH}$	Output Voltage High	$I_{OH} = -2.0\text{mA}$	2.4	3.2					V
		$I_{OH} = -6.5\text{mA}$				2.4	3.2		V

## AC Electrical Characteristics (With Standard Load and Operating Conditions)

Symbol	Parameter	JEDEC Symbol	DM77S321/421			DM87S321/421			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
$T_{AA}$	Address Access Time	TAVQV		40	65		40	55	ns
$T_{EA}$	Enable Access Time	TEVQV		20	35		20	30	ns
$T_{ER}$	Enable Recovery Time	TEXQX		20	35		20	30	ns
$T_{ZX}$	Output Enable Time	TEVQX		20	35		20	30	ns
$T_{XZ}$	Output Disable Time	TEXQZ		20	35		20	30	ns

## AC Electrical Characteristics (With Standard Load and Operating Conditions)

Symbol	Parameter	JEDEC Symbol	DM77S321A/421A			DM87S321A/421A			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
$T_{AA}$	Address Access Time	TAVQV		35			35		ns
$T_{EA}$	Enable Access Time	TEVQV		20			20		ns
$T_{ER}$	Enable Recovery Time	TEXQX		20			20		ns
$T_{ZX}$	Output Enable Time	TEVQX		20			20		ns
$T_{XZ}$	Output Disable Time	TEXQZ		20			20		ns

**Note 3:** These limits apply over the entire operating range unless stated otherwise. All typical values are for  $V_{CC} = 5.0\text{V}$  and  $T_A = 25^\circ\text{C}$ .

**Note 4:** During  $I_{OS}$  measurement, only one output at a time should be grounded. Permanent damage may otherwise result.



## Non-Registered PROM Programming Procedure

National Schottky PROMs are shipped from the factory with all fuses intact. As a result, the outputs will be low (logical "0") for all addresses. To generate high (logical "1") levels at the outputs, the device must be programmed. Information regarding commercially available programming equipment may be obtained from National. If it is desired to build your own programmer, the following conditions must be observed:

1. Programming should be attempted only at ambient temperatures between 15 and 30 degrees Celsius.
2. Address and enable inputs must be driven with TTL logic levels during programming and verification.
3. Programming will occur at the selected address when  $V_{CC}$  is at 10.5 volts, and at the selected bit location when the output pin, representing that bit, is at 10.5 volts, and the device is subsequently enabled. To achieve these conditions in the appropriate sequence, the following procedure must be followed:
  - a) Select the desired word by applying high or low levels to the appropriate address inputs. Disable the device by applying a high level to one or more "active low" chip enable inputs.
  - b) Increase  $V_{CC}$  from nominal to 10.5 volts (plus or minus 0.5V) with a slew rate between 1.0 and 10.0V/ $\mu$ s. Since  $V_{CC}$  is the source of the current required to program the fuse as well as the  $I_{CC}$  for the device at the programming voltage, it must be capable of supplying 750mA at 11.0 volts.
  - c) Select the output where a logical high is desired by raising that output voltage to 10.5 volts (plus or minus 0.5V). Limit the slew rate from 1.0 to 10.0V/ $\mu$ s. This voltage change may occur simultaneously with the increase in  $V_{CC}$ , but must not precede it. It is critical that only one output at a time be programmed since the internal circuits can only supply programming current to one bit at a time. Outputs not being programmed must be left open or connected to a high impedance source of 20k $\Omega$  minimum (Remember that the outputs of the device are disabled at this time).
  - d) Enable the device by taking the chip enable(s) to a low level. This is done with a pulse of 10 $\mu$ s. The 10 $\mu$ s duration refers to the time that the circuit (device) is enabled. Normal input levels are used and rise and fall times are not critical.
  - e) Verify that the bit has been programmed by first removing the programming voltage from the output and then reducing  $V_{CC}$  to 4.0 volts (plus or minus 0.2V). Verification at a  $V_{CC}$  level of 4.0 volts will guarantee proper output states over the  $V_{CC}$  and temperature range of the programmed part. The device must be enabled to sense the state of the outputs. During verification, the loading of the output must be within specified  $I_{OL}$  and  $I_{OH}$  limits. Steps b, c, and d must be repeated up to 10 times or until verification that the bit has been programmed.
  - f) Following verification, apply five additional programming pulses to the bit being programmed. The programming procedure is now complete for the selected bit.
  - g) Repeat steps a through f for each bit to be programmed to a high level. If the procedure is performed on an automatic programmer, the duty cycle of  $V_{CC}$  at the programming voltage must be limited to a maximum of 25%. This is necessary to minimize device junction temperatures. After all selected bits are programmed, the entire contents of the memory should be verified.

Note: Since only an enabled device is programmed, it is possible to program these parts at the board level if all programming parameters are complied with.

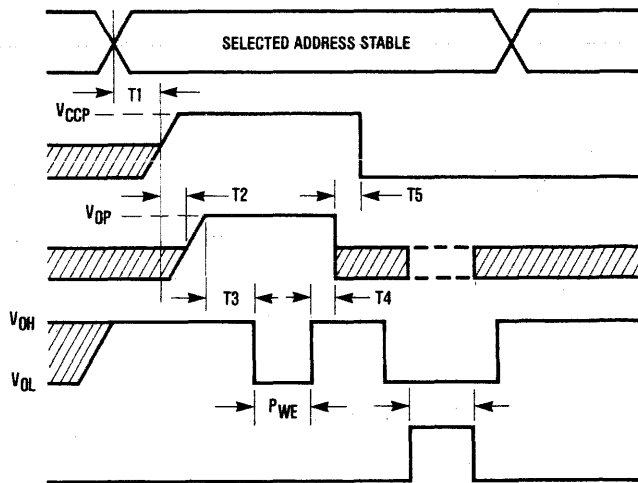
**Programming Parameters** Do not test or you may program the device

Sym	Parameters	Conditions	Min	Recommended Value	Max	Units
V <sub>CCP</sub>	Required V <sub>CC</sub> for Programming		10.0	10.5	11.0	V
I <sub>CCP</sub>	I <sub>CC</sub> During Programming	V <sub>CC</sub> = 11V			750	mA
V <sub>OP</sub>	Required Output Voltage for Programming		10.0	10.5	11.0	V
I <sub>OP</sub>	Output Current while Programming	V <sub>OUT</sub> = 11V			20	mA
I <sub>RR</sub>	Rate of Voltage Change of V <sub>CC</sub> or Output		1.0		10.0	V/μs
P <sub>WE</sub>	Programming Pulse Width (Enabled)		9	10	11	μs
V <sub>CCV</sub>	Required V <sub>CC</sub> for Verification		3.8	4.0	4.2	V
M <sub>DC</sub>	Maximum Duty Cycle for V <sub>CC</sub> at V <sub>CCP</sub>			25	25	%

**Programming Waveforms** Non-Registered PROM

T1 = 100ns min.  
 T2 = 5μs min. T2 may be > 0 if V<sub>CCP</sub> rises at the same rate or faster than (V<sub>OP</sub>)  
 T3 = 100ns min.  
 T4 = 100ns min.  
 T5 = 100ns min.

P<sub>WE</sub> is repeated for 5 additional pulses after verification of V<sub>OH</sub> indicates a bit has been programmed.



## Registered PROM Programming Procedure

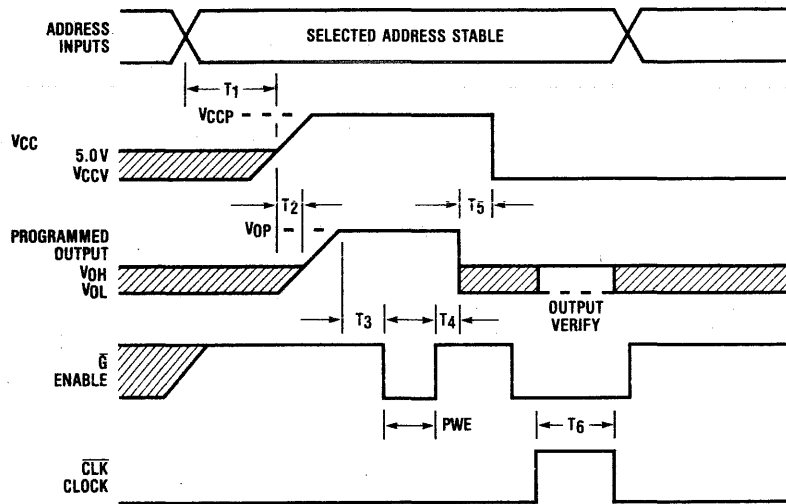
National Schottky PROMs are shipped from the factory with all fuses intact. As a result, the outputs will be low (logical '0') for all addresses. To generate high (logical '1') levels at the outputs, the device must be programmed. Information regarding commercially available programming equipment may be obtained from National. If it is desired to build your own programmer, the following conditions must be observed.

1. Programming should be attempted only at ambient temperatures between 15°C and 30°C.
2. Address and Enable inputs must be driven with TTL logic levels during programming and verification.
3. Programming will occur at the selected address when  $V_{CC}$  is at 10.5V, and at the selected bit location when the output pin, representing that bit, is at 10.5V, and the device is subsequently enabled. To achieve these conditions in the appropriate sequence, the following procedure must be followed:
  - a) Select the desired word by applying high or low levels to the appropriate address inputs. Disable the device by applying a high level to asynchronous chip Enable input  $\overline{G}$ .  $\overline{GS}$  is held low during the entire programming time.
  - b) Increase  $V_{CC}$  from nominal to 10.5V ( $\pm 0.5V$ ) with a slew rate between 1.0 and 10V/ $\mu$ s. Since  $V_{CC}$  is the source of the current required to program the fuse as well as the  $I_{CC}$  for the device at the programming voltage, it must be capable of supplying 750mA at 11V.
  - c) Select the output where a logical high is desired by raising that output voltage to 10.5V ( $\pm 0.5V$ ). Limit the slew rate from 1.0 to 10V/ $\mu$ s. This voltage change may occur simultaneously with the increase in  $V_{CC}$ , but must not precede it. It is critical that only one output at a time be programmed since the internal circuits can only supply programming current to one bit at a time. Outputs not being programmed must be left open or connected to a high impedance source of 20k $\Omega$  minimum. (Remember that the outputs of the device are disabled at this time.)
  - d) Enable the device by taking the chip enable ( $\overline{G}$ ) to a low level. This is done with a pulse of 10 $\mu$ s. The 10 $\mu$ s duration refers to the time that the circuit (device) is enabled. Normal input levels are used and rise and fall times are not critical.
  - e) Verify that the bit has been programmed by first removing the programming voltage from the output and then reducing  $V_{CC}$  to 4.0V ( $\pm 0.2V$ ). Verification at a  $V_{CC}$  level of 4.0V will guarantee proper output states over the  $V_{CC}$  and temperature range of the programmed part. Each data verification must be preceded by a positive going (low to high) clock edge to load the data from the array into the output register. The device must be enabled to sense the state of the outputs. During verification, the loading of the output must be within specified  $I_{OL}$  and  $I_{OH}$  limits. Steps b, c, and d must be repeated up to 10 times or until verification that the bit has been programmed.
  - f) The initialize word is programmed by setting  $\overline{INIT}$  input to a logic low and programming the initialize word output by output in the same manner as any other address. This can be accomplished by inverting the A9 address input from the PROM programmer and applying it to the  $\overline{INIT}$  input. Using this method, the initialize word will program at address 512.
  - g) Following verification, apply five additional programming pulses to the bit being programmed. The programming procedure is now complete for the selected bit.
  - h) Repeat steps a through f for each bit to be programmed to a high level. If the procedure is performed on an automatic programmer, the duty cycle of  $V_{CC}$  at the programming voltage must be limited to a maximum of 25%. This is necessary to minimize device junction temperatures. After all selected bits are programmed, the entire contents of the memory should be verified.

**Programming Parameters** Do not test or you may program the device

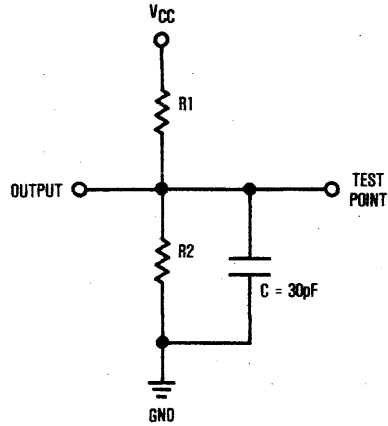
Sym	Parameters	Conditions	Min	Recommended Value	Max	Units
V <sub>CCP</sub>	Required V <sub>CC</sub> for Programming		10.0	10.5	11.0	V
I <sub>CCP</sub>	I <sub>CC</sub> During Programming	V <sub>CC</sub> = 11V			750	mA
V <sub>OP</sub>	Required Output Voltage for Programming		10.0	10.5	11.0	V
I <sub>OP</sub>	Output Current while Programming	V <sub>OUT</sub> = 11V			20	mA
I <sub>RR</sub>	Rate of Voltage Change of V <sub>CC</sub> or Output		1.0		10.0	V/μs
P <sub>WE</sub>	Programming Pulse Width (Enabled)		9	10	11	μs
V <sub>CCV</sub>	Required V <sub>CC</sub> for Verification		3.8	4.0	4.2	V
M <sub>DC</sub>	Maximum Duty Cycle for V <sub>CC</sub> at V <sub>CCP</sub>			25	25	%

**Programming Waveforms** Registered PROM



T<sub>1</sub> = 100 ns MIN.  
 T<sub>2</sub> = 5 μs MIN. (T<sub>2</sub> MAY BE > 0 IF V<sub>CCP</sub> RISES AT THE SAME RATE OR FASTER THAN V<sub>OP</sub>.)  
 T<sub>3</sub> = 100 ns MIN.  
 T<sub>4</sub> = 100 ns MIN.  
 T<sub>5</sub> = 100 ns MIN.  
 T<sub>6</sub> = 50 ns MIN.

Standard Test Load



\* Device input waveform characteristics are:  
 Repetition rate = 1MHz  
 Source impedance = 50Ω  
 Rise and Fall times = 2.5ns max.  
 (1.0 to 2.0 volt levels)

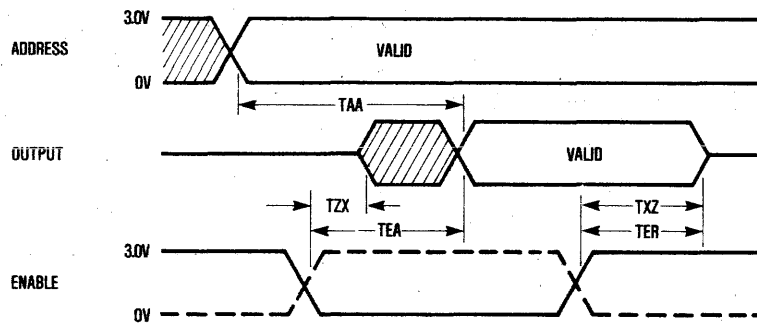
\* TAA is measured with stable enable inputs.

\* TEA and TER are measured from the 1.5 volt level on inputs and outputs with all address and enable inputs stable at applicable levels.

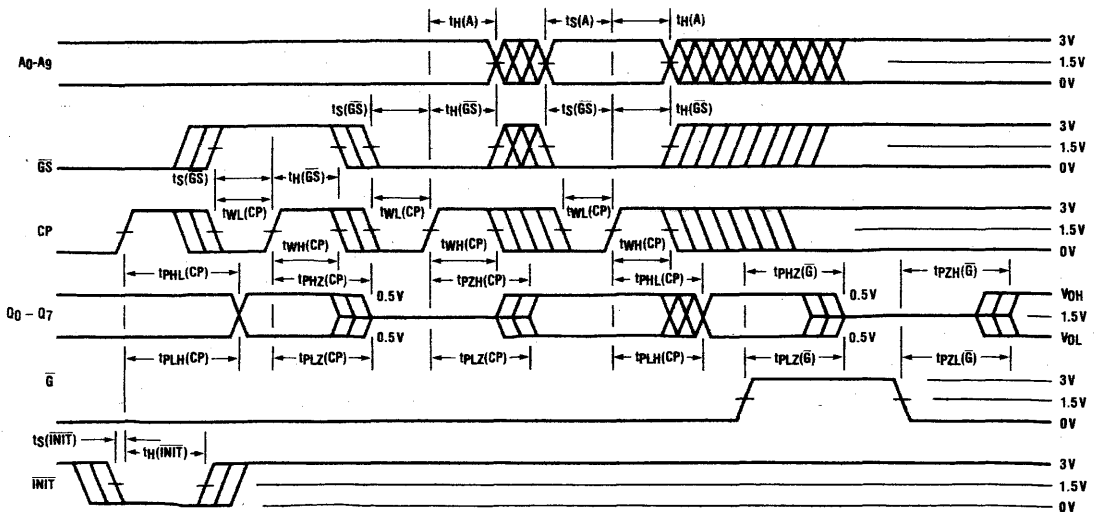
\* For  $I_{OL} = 16mA$ ,  $R1 = 300\Omega$  and  $R2 = 600\Omega$   
 for  $I_{OL} = 12mA$ ,  $R1 = 400\Omega$  and  $R2 = 800\Omega$ .

\* "C" includes scope and jig capacitance.

Switching Time Waveforms Non-Registered PROM



Switching Waveforms Registered PROM



Key To Timing Diagram

WAVEFORM	INPUTS	OUTPUTS	WAVEFORM	INPUTS	OUTPUTS
	MUST BE STEADY	WILL BE STEADY		DON'T CARE: ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
	MAY CHANGE FROM H TO L	WILL BE CHANGING FROM H TO L		DOES NOT APPLY	CENTER LINE IS HIGH IMPEDANCE "OFF" STATE
	MAY CHANGE FROM L TO H	WILL BE CHANGING FROM L TO H			

## Approved Programmers for NSC PROMs

MANUFACTURER	SYSTEM #
DATA I/O	5/17/19/29A
PRO-LOG	M910, M980
KONTRON	MPP80S
STAG	PPX
AIM	RP400
DIGELEC	UP8G3
STARPLEX™	

## Quality Enhancement Programs For Bipolar Memory

A+ PROGRAM*			B+ PROGRAM		
Test	Condition	Guaranteed LOT AQL 5	Test	Condition	Guaranteed LOT AQL 5
D.C. Parametric And Functionality	25° C	0.05	D.C. Parametric And Functionality	25° C	0.05
	Each Temperature Extreme	0.05		Each Temperature Extreme	0.05
A.C. Parametric	25° C	0.4	A.C. Parametric	25° C	0.4
Mechanical	Critical	0.01	Mechanical	Critical	0.01
	Major	0.28		Major	0.28
Seal Tests Hermetic	Fine Leak (5 x 10 <sup>-8</sup> )	0.4	Seal Tests Hermetic	Fine Leak (5 x 10 <sup>-8</sup> )	0.4
	Gross	0.4		Gross	0.4

\*Includes 160 hours of burn-in at 125° C.

 **National Semiconductor**

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# NMC9306/COP494 256-Bit Serial Electrically Erasable Programmable Memory

## General Description

The NMC9306/COP494 is a 256-bit non-volatile sequential access memory fabricated using advanced floating gate N-channel E<sup>2</sup>PROM technology. It is a peripheral memory designed for data storage and/or timing and is accessed via the simple MICROWIRE™ serial interface. The device contains 256 bits of read/write memory divided into 16 registers of 16 bits each. Each register can be serially read or written by a COP400 series controller. Written information is stored in a floating gate cell with at least 10 years data retention and can be updated by an erase-write cycle. The NMC9306/COP494 has been designed to meet applications requiring up to 1 × 10<sup>4</sup> erase/write cycles per register. A power down mode reduces power consumption by 70 percent.

## Features

- Low cost
- Single supply operation (5V ± 10%)
- TTL compatible
- 16 × 16 serial read/write memory
- MICROWIRE compatible serial I/O
- Compatible with COP400 processors
- Low standby power
- Non-volatile erase and write
- Reliable floating gate technology

## Block and Connection Diagrams

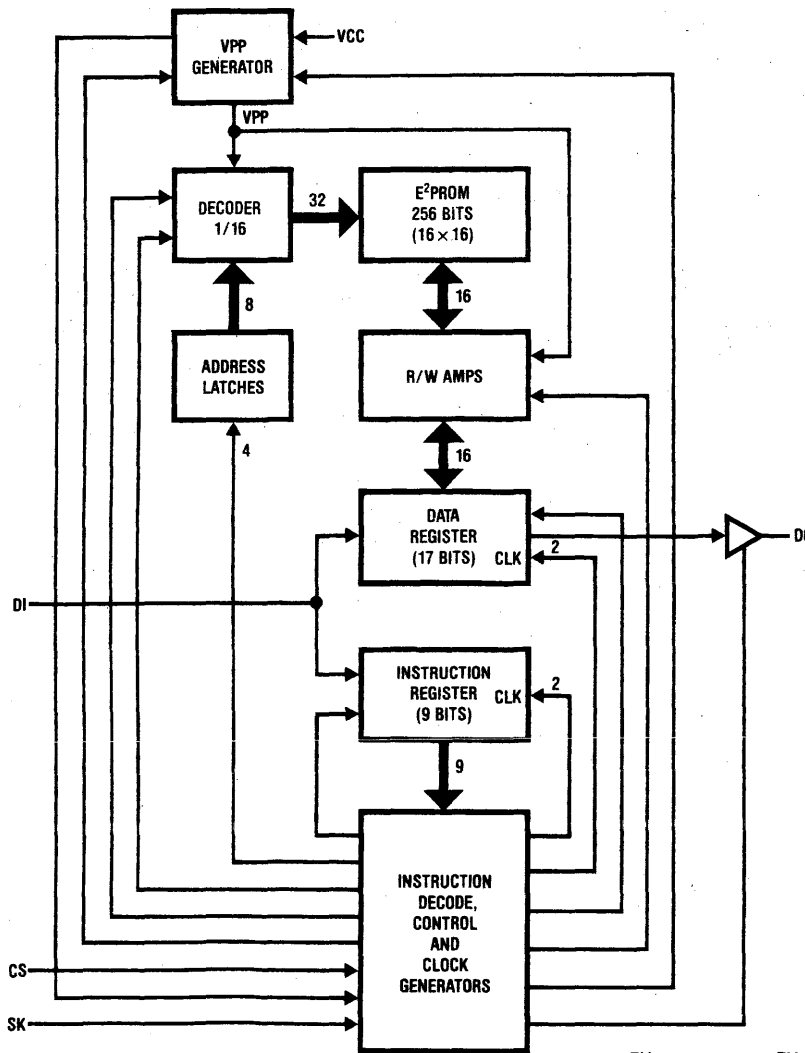
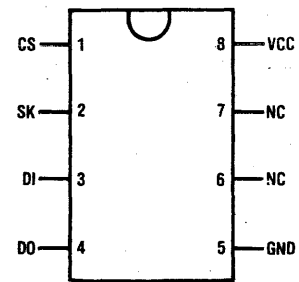


FIGURE 1

COP<sup>TM</sup> and MICROWIRE<sup>TM</sup> are trademarks of National Semiconductor Corp. TRI-STATE<sup>®</sup> is a registered trademark of National Semiconductor Corp.

Dual-In-Line Package



TOP VIEW  
FIGURE 2

### Pin Names

- CS Chip Select
- SK Serial Data Clock
- DI Serial Data Input
- DO Serial Data Output
- VCC Power Supply
- GND Ground

# NMC9346/COP495 1024-Bit Serial Electrically Erasable Programmable Memory (5V Only)

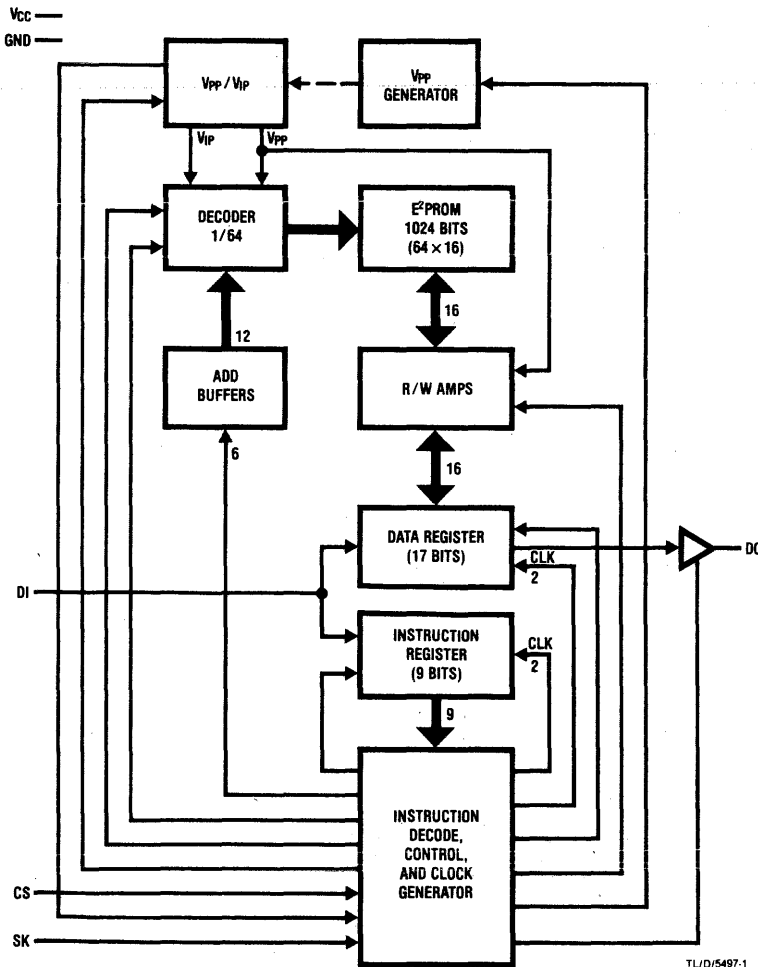
## General Description

The NMC9346/COP495 is a 1024-bit non-volatile, sequential E<sup>2</sup>PROM, fabricated using advanced N-channel E<sup>2</sup>PROM technology. It is an external memory with the 1024 bits of read/write memory divided into 64 registers of 16 bits each. Each register can be serially read or written by a COP400 controller, or a standard microprocessor. Written information is stored in a floating gate cell until updated by an erase and write cycle. The NMC9346/COP495 has been designed for applications requiring up to 10<sup>4</sup> erase/write cycles per register. A power-down mode is provided by CS to reduce power consumption by 75 percent.

## Features

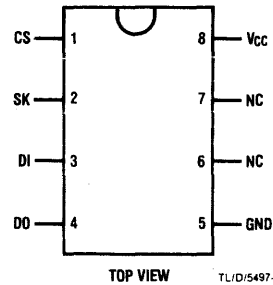
- Low cost
- Single supply read/write/erase operations (5V ± 10%)
- TTL compatible
- 64 × 16 serial read/write memory
- MICROWIRE™ compatible serial I/O
- Simple interfacing
- Low standby power
- Non-volatile erase and write
- Reliable floating gate technology
- Self-timed programming cycle
- Device status signal during programming

## Block Diagram Serial E<sup>2</sup>PROM



## Connection Diagram

### Dual-In-Line Package



### Pin Names

CS	Chip Select
SK	Serial Data Clock
DI	Serial Data Input
DO	Serial Data Output
V <sub>CC</sub>	Power Supply
GND	Ground
NC	Not Connected

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TRI-STATE® is a registered trademark of National Semiconductor Corp.



## NMC9716 16k (2k × 8) Electrically Erasable PROM

Max Access/Current	NMC9716-25	NMC9716-35	NMC9716-45
Max Access Time (ns)	250	350	450
Max Active Current (mA)	110	110	110
Max Standby Current (mA)	50	50	50

### General Description

The NMC9716 is a 16,384-bit electrically erasable and programmable read-only memory (E<sup>2</sup>PROM) fabricated using National's high speed, low power, N-channel double silicon gate technology. The electrical erase/write capability of the NMC9716 makes it ideal for a wide variety of applications requiring in-system, non-volatile erase and write.

The NMC9716 is pin and functionally compatible with the NMC2816 E<sup>2</sup>PROM, with the added system feature of erasing/writing with a 5V TTL pulse on chip enable ( $\overline{CE}$ ), while the VPP is held at 21V. The erase/write cycle is very similar to the industry standard 2716 EPROM programming cycle.

The device operates from a 5V power supply in the read mode and, with its very fast read access speed, is compatible with high performance microprocessors.

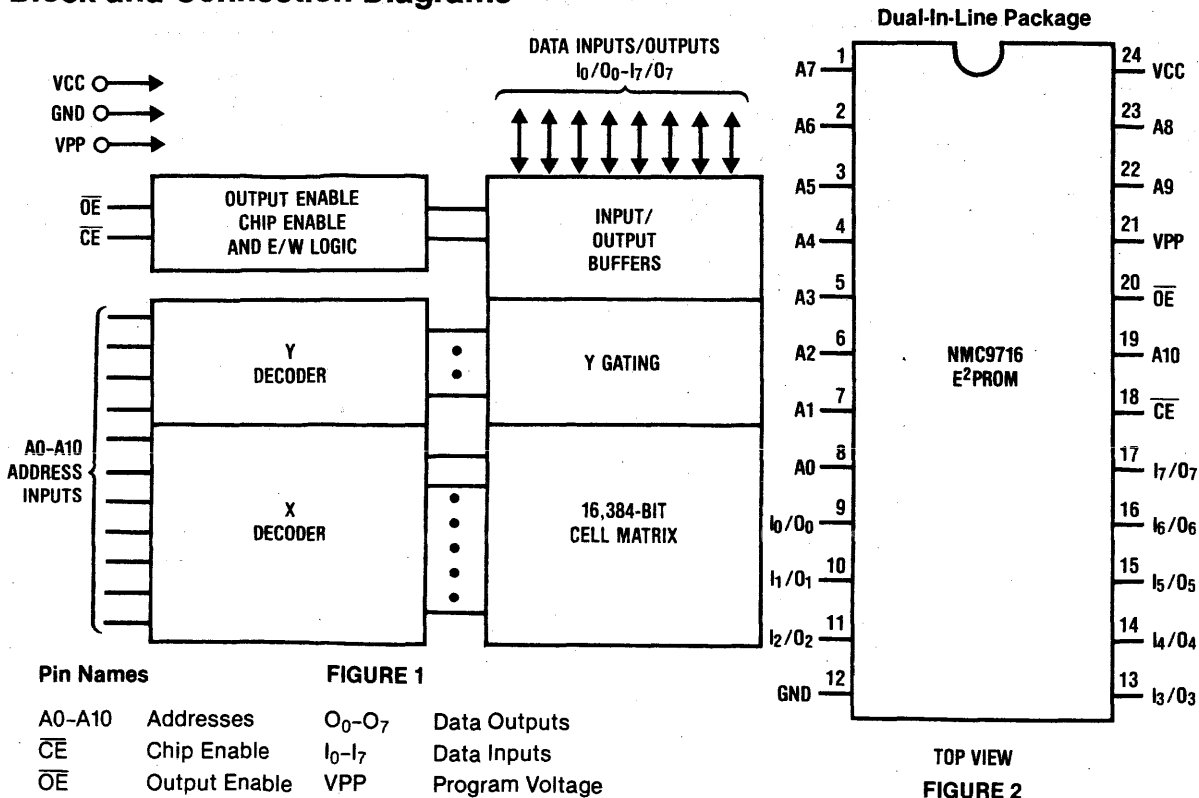
The NMC9716 is deselected when  $\overline{CE}$  input is high and is automatically placed in the standby mode. This mode provides a 52% reduction in power with no increase in access time. The NMC9716 also has an output enable control to eliminate bus contention in a system environment.

The NMC9716 can be easily erased and reprogrammed in a byte-by-byte mode and the entire memory array can be erased with a single programming pulse in the chip erase mode. Byte erase is identical to byte write, with all inputs at logic one (TTL high).

### Features

- Erase/write with a 5V TTL pulse or a 21V pulse
- Pin and functionally compatible with the NMC2816
- No rise time restriction on erase/write pulse
- 2048 × 8 organization
- Conforms to JEDEC byte-wide family standard
- Microprocessor compatible architecture
- Single byte erase/write capability
- 10 ms byte erase/write time
- 10 ms chip erase mode
- Low power dissipation
  - 610 mW max (active power ICC + IPP)
  - 295 mW max (standby power ICC + IPP)

### Block and Connection Diagrams



# NMC9817 16,384-Bit (2k x 8) E<sup>2</sup>PROM

## General Description

The NMC9817 is a fast 5V-only E<sup>2</sup>PROM which offers many desired application features, making it ideally suited for efficiency and ease in system design. The added features on the NMC9817 include: 5V-only operation provided by an on-chip V<sub>PP</sub> generation during erase-write; address and data latches to reduce part count and to free the microprocessor while the chip is busy during erase-write; 'ready' line indicator to indicate status of chip to the microprocessor and automatic erase before byte-write. It can meet applications requiring up to 10<sup>4</sup> write cycles per byte. The NMC9817 is a product of National's advanced E<sup>2</sup>PROM stepper technology and uses the powerful XMOS™ process for reliable, non-volatile data storage.

The NMC9817 sharply minimizes the interfacing hardware logic and firmware required to perform data writes. The device has complete self-timing which leaves the processor free to perform other tasks until the NMC9817 signals 'ready'. With an automatic erase before write, the user benefits by saving an erase command contributing to efficient usage of system processing time. On-chip address and data latching further enhances system performance.

The NMC9817's very fast read access times make it compatible with high performance microprocessor applications. It uses the proven two line control architecture which eliminates bus contention in a system environment. Combining these features with the NMC9817's 'ready' signal makes the device an extremely powerful, yet simple to use, E<sup>2</sup> memory.

The density, and level of integrated control, make the NMC9817 suitable for users requiring no hardware overhead, high system performance, minimal board space and design ease. Designing with and using the NMC9817 is extremely cost effective as the required high voltage and interfacing hardware required for other E<sup>2</sup>PROM devices has been eliminated by 5V-only operation and on-chip latches. See *Figures 1, 2, and 3* for the NMC9817 block diagram, pinout, and simple interface requirements.

## Features

- Single 5V supply (eliminates an external 21V V<sub>PP</sub>)
- Self-timed byte-write with auto erase
- No external capacitor or pulse shaping circuits
- On-chip address and data latches
- Two line output control
- TRI-STATE® outputs
- RDY pin indicator
- Fast byte-writing  
Write cycle (2 ms typical)  
E/W cycle (4 ms typical)
- Very fast access times  
NMC9817-20—200 ns  
NMC9817-25—250 ns  
NMC9817-35—350 ns
- Direct microprocessor interface capability
- No support components needed
- Reliable E<sup>2</sup>PROM XMOS stepper technology

## Block and Connection Diagrams

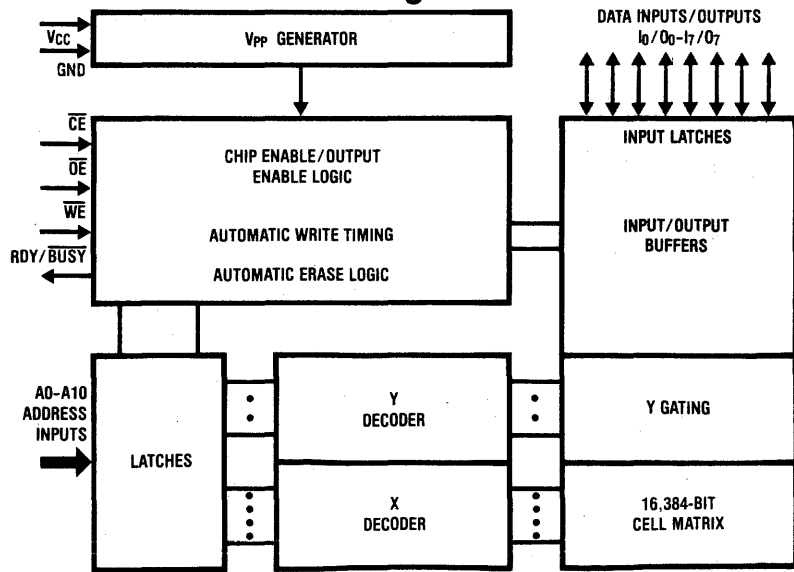
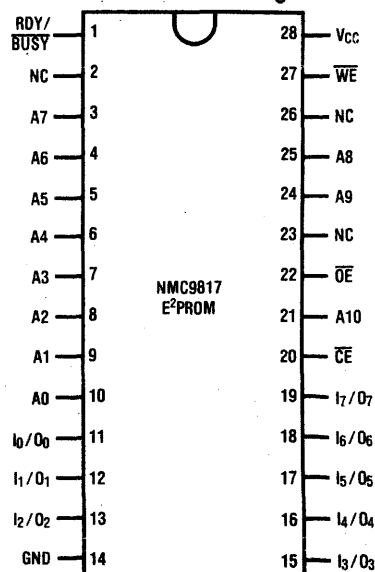


FIGURE 1

TL/D/5041-1

## Dual-In-Line Package



TOP VIEW

FIGURE 2

TL/D/5041-2

Pin Name	Symbol	Function
A0-A10	Addresses	Addresses
CE	Chip Enable	RDY/BUSY
OE	Output Enable	NC
O0-O7	Data Outputs	Data Inputs
		Device Ready/Busy
		No Connect

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## HS2649 65,536 x 9 Bit Dynamic Random-Access Memory Module

### General Description

The HS2649 is a high speed high density 589,824 bit, dynamic random-access memory organized into 65,536 words of 9 bits each (8 bit data plus parity). Single +5 volt power supply operation with internal decoupling capacitors allows for very compact layout of large memory systems.

The HS2649 is in a 30 pin single-in-line package (SIP) which uses a minimum of board space. This SIP package features ceramic construction, high strength, and good thermal conductivity.

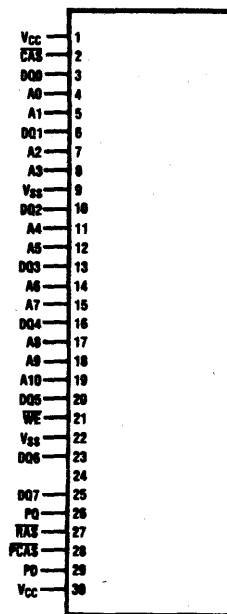
### Applications

- Mainframe memory
- Buffer memory
- Peripheral storage
- Transportable computers

### Features

- 65536×9 Bit Dynamic Ram
- SIP package
- 200 ns access time
- 335 ns cycle time
- Single +5 volt supply  $\pm 10\%$
- Common I/O except parity bit
- All inputs TTL compatible
- 256 refresh cycles/4 ms
- Power 2.8W active max, 180 mW standby max.

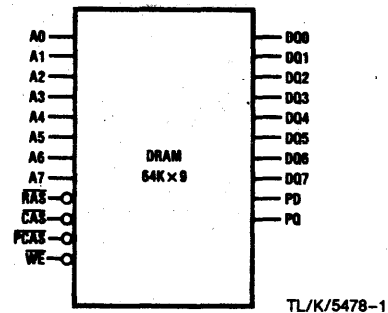
### Connection Diagram\*



TL/K/5478-2

\*A8, 9 & 10 are designated for reference for future products, and are no connections on this part.

### Logic Symbol



### Pin Designations

A0 thru A7	Address Inputs
DQ0 thru DQ7	Data Inputs/Outputs
PD	Parity Input
PQ	Parity Output
$\overline{\text{RAS}}$	Row Address Strobe
$\overline{\text{CAS}}$	Column Address Strobe
$\overline{\text{PCAS}}$	Parity Column Address Strobe
$\overline{\text{WE}}$	Write Enable
V <sub>CC</sub>	+5 Volt Supply
V <sub>SS</sub>	Ground



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**Vol. II, page 4425**

**DESCRIPTION** The NEC μPD416 is a 16384 words by 1 bit Dynamic MOS RAM. It is designed for memory applications where very low cost and large bit storage are important design objectives.

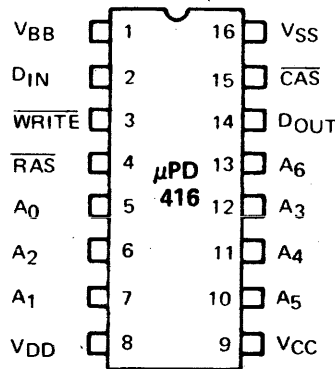
The μPD416 is fabricated using a double-poly-layer N channel silicon gate process which affords high storage cell density and high performance. The use of dynamic circuitry throughout, including the sense amplifiers, assures minimal power dissipation.

Multiplexed address inputs permit the μPD416 to be packaged in the standard 16 pin dual-in-line package. The 16 pin package provides the highest system bit densities and is available in either ceramic or plastic. Noncritical clock timing requirements allow use of the multiplexing technique while maintaining high performance.

- FEATURES**
- 16384 Words x 1 Bit Organization
  - High Memory Density – 16 Pin Ceramic and Plastic Packages
  - Multiplexed Address Inputs
  - Standard Power Supplies +12V, -5V, +5V
  - Low Power Dissipation; 462 mW Active (MAX), 20 mW Standby (MAX)
  - Output Data Controlled by  $\overline{\text{CAS}}$  and Unlatched at End of Cycle
  - Read-Modify-Write,  $\overline{\text{RAS}}$ -only Refresh, and Page Mode Capability
  - All Inputs TTL Compatible, and Low Capacitance
  - 128 Refresh Cycles
  - 5 Performance Ranges:

	ACCESS TIME	R/W CYCLE	RMW CYCLE
μPD416-2	200 ns	375 ns	375 ns
μPD416-3	150 ns	320 ns	320 ns
μPD416-5	120 ns	320 ns	320 ns

**PIN CONFIGURATION**



A <sub>0</sub> -A <sub>6</sub>	Address Inputs
$\overline{\text{CAS}}$	Column Address Strobe
D <sub>IN</sub>	Data In
D <sub>OUT</sub>	Data Out
$\overline{\text{RAS}}$	Row Address Strobe
WRITE	Read/Write
V <sub>BB</sub>	Power (-5V)
V <sub>CC</sub>	Power (+5V)
V <sub>DD</sub>	Power (+12V)
V <sub>SS</sub>	Ground

Rev/4

DC CHARACTERISTICS

$T_a = 0^\circ\text{C to } +70^\circ\text{C}$  ①,  $V_{DD} = +12\text{V} \pm 10\%$ ,  $V_{CC} = +5\text{V} \pm 10\%$ ,  $V_{BB} = -5\text{V} \pm 10\%$ ,  $V_{SS} = 0\text{V}$

PARAMETER	SYMBOL	LIMITS			UNIT	TEST CONDITIONS
		MIN	TYP	MAX		
Supply Voltage	$V_{DD}$	10.8	12.0	13.2	V	②
Supply Voltage	$V_{CC}$	4.5	5.0	5.5	V	② ③
Supply Voltage	$V_{SS}$	0	0	0	V	②
Supply Voltage	$V_{BB}$	-4.5	-5.0	-5.5	V	②
Input High (Logic 1) Voltage, RAS, CAS, WRITE	$V_{IHC}$	2.7		7.0	V	②
Input High (Logic 1) Voltage, all inputs except RAS, CAS, WRITE	$V_{IH}$	2.4		7.0	V	②
Input Low (Logic 0) Voltage, all inputs	$V_{IL}$	1.0		0.8	V	②
Operating $V_{DD}$ Current	$I_{DD1}$			35	mA	RAS, CAS cycling, $t_{RC} = t_{RC} \text{ Min.}$ ④
Standby $V_{DD}$ Current	$I_{DD2}$			1.5	mA	RAS, $V_{IHC}$ , DOUT High Impedance
Refresh $V_{DD}$ Current	All Speeds except μPD416-5	$I_{DD3}$		25	mA	RAS cycling, CAS, $V_{IHC}$ ; $t_{RC} = 375 \text{ ns}$ ④
	μPD416-5	$I_{DD3}$		27	mA	
Page Mode $V_{DD}$ Current	$I_{DD4}$			27	mA	RAS = $V_{IL}$ , CAS cycling; $t_{PC} = 225 \text{ ns}$ ④
Operating $V_{CC}$ Current	$I_{CC1}$				μA	RAS, CAS cycling, $t_{RC} = 375 \text{ ns}$ ⑤
Standby $V_{CC}$ Current	$I_{CC2}$	10		10	μA	RAS, $V_{IHC}$ , DOUT High Impedance
Refresh $V_{CC}$ Current	$I_{CC3}$	10		10	μA	RAS cycling, CAS, $V_{IHC}$ , $t_{RC} = 375 \text{ ns}$
Page Mode $V_{CC}$ Current	$I_{CC4}$				μA	RAS, $V_{IL}$ , CAS cycling; $t_{PC} = 225 \text{ ns}$ ⑤
Operating $V_{BB}$ Current	$I_{BB1}$			200	μA	RAS, CAS cycling, $t_{RC} = 375 \text{ ns}$
Standby $V_{BB}$ Current	$I_{BB2}$			100	μA	RAS, $V_{IHC}$ , DOUT High Impedance
Refresh $V_{BB}$ Current	$I_{BB3}$			200	μA	RAS cycling, CAS, $V_{IHC}$ , $t_{RC} = 375 \text{ ns}$
Page Mode $V_{BB}$ Current	$I_{BB4}$			200	μA	RAS, $V_{IL}$ , CAS cycling; $t_{PC} = 225 \text{ ns}$
Input Leakage (any input)	$I_{I(L)}$	-10		10	μA	$V_{BB} = -5\text{V}, 0\text{V}$ , $V_{IN} \leq +7\text{V}$ , all other pins not under test = 0V
Output Leakage	$I_{O(L)}$	-10		10	μA	DOUT is disabled, $0\text{V} \leq V_{OUT} \leq +5.5\text{V}$
Output High Voltage (Logic 1)	$V_{OH}$	2.4			V	$I_{OUT} = 5 \text{ mA}$ ③
Output Low Voltage (Logic 0)	$V_{OL}$			0.4	V	$I_{OUT} = 4.2 \text{ mA}$

Notes: ①  $T_a$  is specified here for operation at frequencies to  $t_{RC} > t_{RC} \text{ (min)}$ . Operation at higher cycle rates with reduced ambient temperatures and high power dissipation is permissible, however, provided AC operating parameters are met. See Figure 1 for derating curve.

② All voltages referenced to  $V_{SS}$ .

③ Output voltage will swing from  $V_{SS}$  to  $V_{CC}$  when activated with no current loading. For purposes of maintaining data in standby mode,  $V_{CC}$  may be reduced to  $V_{SS}$  without affecting refresh operations or data retention. However, the  $V_{OH} \text{ (min)}$  specification is not guaranteed in this mode.

④  $I_{DD1}$ ,  $I_{DD3}$ , and  $I_{DD4}$  depend on cycle rate. See Figures 2, 3 and 4 for  $I_{DD}$  limits at other cycle rates.

⑤  $I_{CC1}$  and  $I_{CC4}$  depend upon output loading. During readout of high level data  $V_{CC}$  is connected through a low impedance (135Ω typ) to data out. At all other times  $I_{CC}$  consists of leakage currents only.

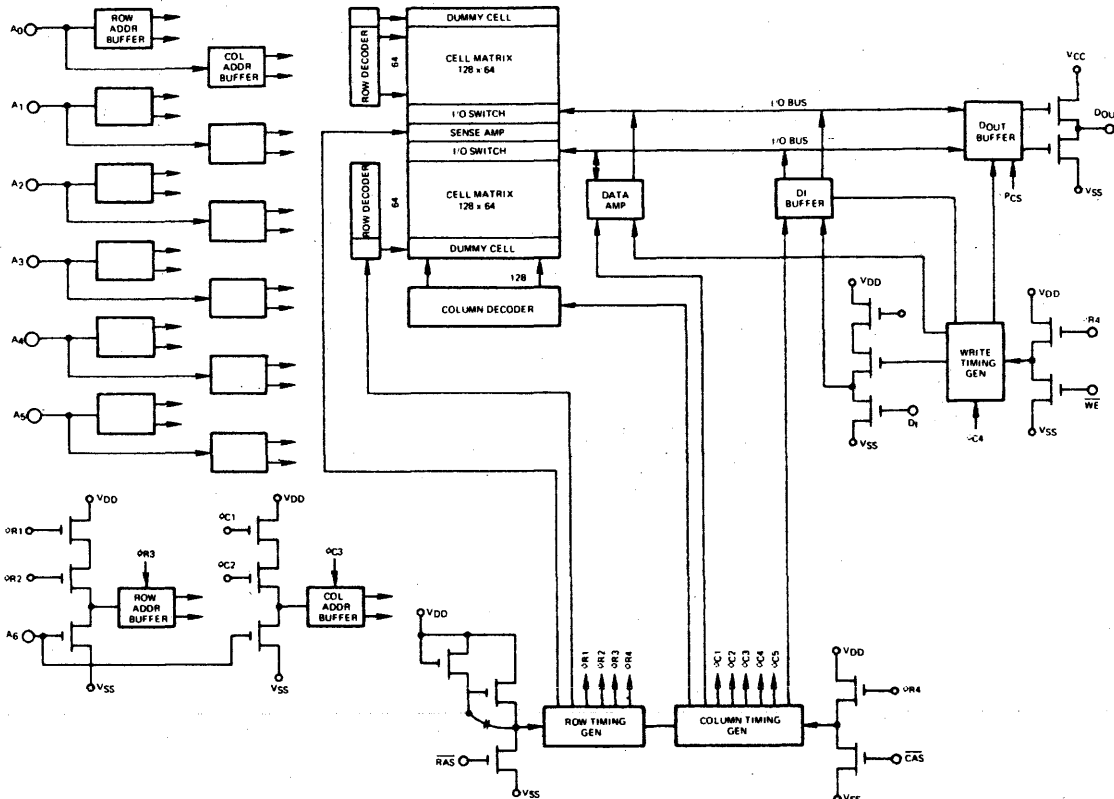
AC  
CHARACTERISTICS

T<sub>a</sub> = 0°C to +70°C, V<sub>DD</sub> = +12V ± 10%, V<sub>CC</sub> = +5V ± 10%, V<sub>BB</sub> = -5V ± 10%, V<sub>SS</sub> = 0V

PARAMETER	SYMBOL	LIMITS						UNIT	TEST CONDITIONS
		μPD416-2		μPD416-3		μPD416-5			
		MIN	MAX	MIN	MAX	MIN	MAX		
Random read or write cycle time	t <sub>RC</sub>	375		320		320		ns	③
Read-write cycle time	t <sub>RWC</sub>	375		375		320		ns	③
Page mode cycle time	t <sub>PC</sub>	225		170		160		ns	
Access time from RAS	t <sub>RAC</sub>		200		150		120	ns	④ ⑥
Access time from CAS	t <sub>CAC</sub>		135		100		80	ns	⑤ ⑥
Output buffer turn-off delay	t <sub>OFF</sub>	0	50	0	40	0	35	ns	⑦
Transition time (rise and fall)	t <sub>T</sub>	3	50	3	35	3	35	ns	②
RAS precharge time	t <sub>RP</sub>	120		100		100		ns	
RAS pulse width	t <sub>RAS</sub>	200	32,000	150	32,000	120	10,000	ns	
RAS hold time	t <sub>RS</sub>	135		100		80		ns	
CAS pulse width	t <sub>CAS</sub>	135	10,000	100	10,000	80	10,000	ns	
RAS to CAS delay time	t <sub>RCD</sub>	25	65	20	50	15	40	ns	⑧
CAS to RAS precharge time	t <sub>CRP</sub>	-20		-20		0		ns	
Row address set-up time	t <sub>ASR</sub>	0		0		0		ns	
Row address hold time	t <sub>RAH</sub>	25		20		15		ns	
Column address set-up time	t <sub>ASC</sub>	-10		-10		10		ns	
Column address hold time	t <sub>CAH</sub>	55		45		40		ns	
Column address hold time referenced to RAS	t <sub>AR</sub>	120		95		80		ns	
Read command set-up time	t <sub>RCS</sub>	0		0		0		ns	
Read command hold time	t <sub>RCH</sub>	0		0		0		ns	
Write command hold time	t <sub>WCH</sub>	55		45		40		ns	
Write command hold time referenced to RAS	t <sub>WCR</sub>	120		95		80		ns	
Write command pulse width	t <sub>WP</sub>	55		45		40		ns	
Write command to RAS lead time	t <sub>RWL</sub>	70		50		50		ns	
Write command to CAS lead time	t <sub>CWL</sub>	70		50		50		ns	
Data-in set-up time	t <sub>DS</sub>	0		0		0		ns	⑨
Data-in hold time	t <sub>DH</sub>	55		45		40		ns	⑨
Data-in hold time referenced to RAS	t <sub>DHR</sub>	120		95		80		ns	
CAS precharge time (for page mode cycle only)	t <sub>CP</sub>	80		60		60		ns	
Refresh period	t <sub>REF</sub>		2		2		2	ms	
WRITE command set-up time	t <sub>WCS</sub>	20		20		0		ns	⑩
CAS to WRITE delay	t <sub>CWD</sub>	95		70		80		ns	⑩
RAS to WRITE delay	t <sub>RWD</sub>	160		120		120		ns	⑩

- Notes:
- ① AC measurements assume t<sub>T</sub> = 5 ns.
  - ② V<sub>IHC</sub> (min) or V<sub>IHL</sub> (min) and V<sub>IHL</sub> (max) are reference levels for measuring timing of input signals. Also, transition times are measured between V<sub>IHC</sub> or V<sub>IHL</sub> and V<sub>IL</sub>.
  - ③ The specifications for t<sub>RC</sub> (min) and t<sub>RWC</sub> (min) are used only to indicate cycle time at which proper operation over the full temperature range (0°C < T<sub>a</sub> < 70°C) is assured.
  - ④ Assumes that t<sub>RCD</sub> < t<sub>RCD</sub> (max). If t<sub>RCD</sub> is greater than the maximum recommended value shown in this table, t<sub>RAC</sub> will increase by the amount that t<sub>RCD</sub> exceeds the values shown.
  - ⑤ Assumes that t<sub>RCD</sub> > t<sub>RCD</sub> (max).
  - ⑥ Measured with a load equivalent to 2 TTL loads and 100 pF.
  - ⑦ t<sub>OFF</sub> (max) defines the time at which the output achieves the open circuit condition and is not referenced to output voltage levels.
  - ⑧ Operation within the t<sub>RCD</sub> (max) limit ensures that t<sub>RAC</sub> (max) can be met. t<sub>RCD</sub> (max) is specified as a reference point only. If t<sub>RCD</sub> is greater than the specified t<sub>RCD</sub> (max) limit, then access time is controlled exclusively by t<sub>CAC</sub>.
  - ⑨ These parameters are referenced to CAS leading edge in early write cycles and to WRITE leading edge in delayed write or read-modify-write cycles.
  - ⑩ t<sub>WCS</sub>, t<sub>CWD</sub> and t<sub>RWD</sub> are not restrictive operating parameters. They are included in the data sheet as electrical characteristics only. If t<sub>WCS</sub> < t<sub>WCS</sub> (min), the cycle is an early write cycle and the data out pin will remain open circuit (high impedance) > t<sub>RWD</sub> (min), the cycle is a read-write cycle and the data out will contain data read from the selected cell; if neither of the above sets of conditions is satisfied the condition of the data out (at access time) is indeterminate.

**BLOCK  
DIAGRAM**



Operating Temperature .....	0°C to +70°C
Storage Temperature .....	-55°C to +150°C
All Output Voltages ① .....	-0.5 to +20 Volts
All Input Voltages ① .....	-0.5 to +20 Volts
Supply Voltages V <sub>DD</sub> , V <sub>CC</sub> , V <sub>SS</sub> ① .....	-0.5 to +20 Volts
Supply Voltages V <sub>DD</sub> , V <sub>CC</sub> ② .....	-1.0 to +15 Volts
Short Circuit Output Current .....	50 mA
Power Dissipation .....	1 Watt

**ABSOLUTE MAXIMUM RATINGS\***

Notes: ① Relative to V<sub>BB</sub>  
 ② Relative to V<sub>SS</sub>  
 T<sub>a</sub> = 25°C

\*COMMENT: Stress above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

T<sub>a</sub> = 0°C to 70°C, V<sub>DD</sub> = +12V ± 10%, V<sub>BB</sub> = -5V ± 10%, V<sub>CC</sub> = +5V ± 10%, V<sub>SS</sub> = 0V

**CAPACITANCE**

PARAMETER	SYMBOL	LIMITS			UNIT	TEST CONDITIONS
		MIN	TYP	MAX		
Input Capacitance (A <sub>0</sub> -A <sub>6</sub> ), D <sub>IN</sub>	C <sub>I1</sub>		4	5	pF	
Input Capacitance RAS, CAS, WRITE	C <sub>I2</sub>		8	10	pF	
Output Capacitance (D <sub>OUT</sub> )	C <sub>O</sub>		5	7	pF	



**DESCRIPTION** The NEC  $\mu$ PD4164 is a 65,536 words by 1 bit Dynamic N-Channel MOS RAM designed to operate from a single +5V power supply. The negative-voltage substrate bias is internally generated — its operation is both automatic and transparent.

The  $\mu$ PD4164 utilizes a three-poly N-channel silicon gate process which provides high storage cell density, high performance and high reliability.

The  $\mu$ PD4164 uses a single transistor dynamic storage cell and advanced dynamic circuitry throughout, including the 512 sense amplifiers, which assures that power dissipation is minimized. Refresh characteristics have been chosen to maximize yield (low cost to user) while maintaining compatibility between Dynamic RAM generations.

The  $\mu$ PD4164 three-state output is controlled by  $\overline{\text{CAS}}$ , independent of  $\overline{\text{RAS}}$ . After a valid read or read-modify-write cycle, data is held on the output by holding  $\overline{\text{CAS}}$  low. The data out pin is returned to the high impedance state by returning  $\overline{\text{CAS}}$  to a high state. The  $\mu$ PD4164 hidden refresh feature allows  $\overline{\text{CAS}}$  to be held low to maintain output data while  $\overline{\text{RAS}}$  is used to execute  $\overline{\text{RAS}}$  only refresh cycles.

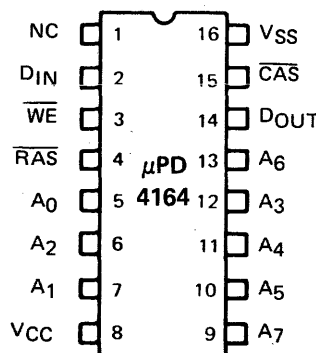
Refreshing is accomplished by performing  $\overline{\text{RAS}}$  only refresh cycles, hidden refresh cycles, or normal read or write cycles on the 128 address combinations of  $A_0$  through  $A_6$  during a 2 ms period.

Multiplexed address inputs permit the  $\mu$ PD4164 to be packaged in the standard 16 pin dual-in-line package. The 16 pin package provides the highest system bit densities and is compatible with widely available automated handling equipment.

**FEATURES**

- High Memory Density
- Multiplexed Address Inputs
- Single +5V Supply
- On Chip Substrate Bias Generator
- Access Time:  $\mu$ PD4164-20 — 200 ns  
 $\mu$ PD4164-15 — 150 ns  
 $\mu$ PD4164-12 — 120 ns
- Read, Write Cycle Time:  $\mu$ PD4164-20 — 335 ns  
 $\mu$ PD4164-15 — 270 ns  
 $\mu$ PD4164-12 — 260 ns
- Low Power Dissipation: 250 mW (Active); 28 mW (Standby)
- Non-Latched Output is Three-State, TTL Compatible
- Read, Write, Read-Write; Read-Modify-Write,  $\overline{\text{RAS}}$  Only Refresh, and Page Mode Capability
- All Inputs TTL Compatible, and Low Input Capacitance
- 128 Refresh Cycles ( $A_0$ - $A_6$  Pins for Refresh Address)
- $\overline{\text{CAS}}$  Controlled Output Allows Hidden Refresh
- Available in Both Ceramic and Plastic 16 Pin Packages

**PIN CONFIGURATION**



**PIN NAMES**

$A_0$ - $A_7$	Address Inputs
$\overline{\text{RAS}}$	Row Address Strobe
$\overline{\text{CAS}}$	Column Address Strobe
$\overline{\text{WE}}$	Write Enable
DIN	Data Input
DOUT	Data Output
VCC	Power Supply (+5V)
VSS	Ground
NC	No Connection

Rev/2

Operating Temperature	0°C to +70°C
Storage Temperature (Ceramic Package)	-55°C to +150°C
(Plastic Package)	-55°C to +125°C
Supply Voltages On Any Pin Except V <sub>CC</sub>	-1 to +7 Volts ①
Supply Voltage V <sub>CC</sub>	-0.5 to +7 Volts ①
Short Circuit Output Current	50 mA
Power Dissipation	1 Watt

**ABSOLUTE MAXIMUM RATINGS\***

Note: ① Relative to V<sub>SS</sub>

T<sub>a</sub> = 25°C

\*COMMENT: Stress above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

T<sub>a</sub> = 0° to 70°C ① ; V<sub>CC</sub> = +5V ± 10%; V<sub>SS</sub> = 0V

**DC CHARACTERISTICS**

PARAMETER	SYMBOL	LIMITS			UNIT	TEST CONDITIONS
		MIN	TYP	MAX		
Supply Voltage	V <sub>CC</sub>	4.5	5.0	5.5	V	All Voltages Referenced to V <sub>SS</sub>
	V <sub>SS</sub>	0	0	0	V	
High Level Input Voltage, (RAS, CAS, WE)	V <sub>IHC</sub>	2.4		5.5	V	
High Level Input Voltage, All Inputs Except RAS, CAS, WE	V <sub>IH</sub>	2.4		5.5	V	
Low Level Input Voltage, All Inputs	V <sub>IL</sub>	-2.0		0.8	V	
Operating Current Average Power Supply Operating Current RAS, CAS Cycling; t <sub>RC</sub> = t <sub>RC</sub> (Min.)	I <sub>CC1</sub>			45	mA	
				50		
				55		
Standby Current Power Supply Standby Current (RAS = V <sub>IHC</sub> , D <sub>OUT</sub> = Hi-Impedance)	I <sub>CC2</sub>			5.0	mA	
Refresh Current Average Power Supply Current, Refresh Mode; RAS Cycling, CAS = V <sub>IHC</sub> , t <sub>RC</sub> = t <sub>RC</sub> (Min.)	I <sub>CC3</sub>			35	mA	②
				40		
				45		
Page Mode Current Average Power Supply Current, Page Mode Operation RAS = V <sub>IL</sub> ; CAS Cycling t <sub>PC</sub> = t <sub>PC</sub> (Min.)	I <sub>CC4</sub>			35	mA	②
				40		
				45		
Input Leakage Current Any Input V <sub>IN</sub> = 0 to +5.5 Volts, All Other Pins Not Under Test = 0V	I <sub>I(L)</sub>	-10		10	μA	
Output Leakage Current D <sub>OUT</sub> is Disabled, V <sub>OUT</sub> = 0 to +5.5 Volts	I <sub>O(L)</sub>	-10		10	μA	
Output Levels High Level Output Voltage (I <sub>O</sub> = 5 mA) Low Level Output Voltage (I <sub>O</sub> = 4.2 mA)	V <sub>OH</sub>	2.4		V <sub>CC</sub>	V	
	V <sub>OL</sub>	0		0.4	V	

- Notes: ① T<sub>a</sub> is specified here for operation at frequencies to t<sub>RC</sub> ≥ t<sub>RC</sub> (min). Operation at higher cycle rates with reduced ambient temperatures and high power dissipation is permissible, however, provided AC operating parameters are met.
- ② I<sub>CC1</sub>, I<sub>CC3</sub> and I<sub>CC4</sub> depend on output loading and cycle rates. Specified rates are obtained with the output open.

MEMORY  
NEC Electronics

AC CHARACTERISTICS

$T_a = 0^\circ$  to  $+70^\circ\text{C}$  ①;  $V_{CC} = +5\text{V} \pm 10\%$ ;  $V_{SS} = 0\text{V}$  ③ ④

PARAMETER	SYMBOL	LIMITS						UNIT	TEST CONDITIONS
		μPD4164-20		μPD4164-15		μPD4164-12			
		MIN	MAX	MIN	MAX	MIN	MAX		
Random Read or Write Cycle Time	$t_{RC}$	330		260		230		ns	⑤
Read Write Cycle Time	$t_{RWC}$	345		280		255		ns	⑤
Page Mode Cycle Time	$t_{PC}$	190		145		130		ns	
Access Time from RAS	$t_{RAC}$		200		150		120	ns	⑥ ⑧
Access Time from CAS	$t_{CAC}$		100		75		60	ns	⑦ ⑧
Output Buffer Turn-Off Delay	$t_{OFF}$	0	50	0	40	0	35	ns	⑨
Transition Time (Rise and Fall)	$t_T$	3	50	3	50	3	35	ns	④
RAS Precharge Time	$t_{RP}$	120		100		90		ns	
RAS Pulse Width	$t_{RAS}$	200	10,000	150	10,000	120	10,000	ns	
RAS Hold Time	$t_{RSH}$	100		75		60		ns	
CAS Pulse Width	$t_{CAS}$	100	10,000	75	10,000	60	10,000	ns	
CAS Hold Time	$t_{CSH}$	200		150		120		ns	
RAS to CAS Delay Time	$t_{RCD}$	30	100	25	75	25	60	ns	⑩
CAS to RAS Precharge Time	$t_{CRP}$	0		0		0		ns	
CAS Precharge Time	$t_{CPN}$	30		25		25		ns	
CAS Precharge Time (For Page Mode Cycle Only)	$t_{CP}$	80		60		60		ns	
RAS Precharge CAS Hold Time	$t_{RPC}$	0		0		0		ns	
Row Address Set-Up Time	$t_{ASR}$	0		0		0		ns	
Row Address Hold Time	$t_{RAH}$	20		15		15		ns	
Column Address Set-Up Time	$t_{ASC}$	0		0		0		ns	
Column Address Hold Time	$t_{CAH}$	30		25		20		ns	
Column Address Hold Time Referenced to RAS	$t_{AR}$	130		100		80		ns	
Read Command Set-Up Time	$t_{RCS}$	0		0		0		ns	
Read Command Hold Time Referenced to RAS	$t_{RRH}$	25		20		20		ns	⑬
Read Command Hold Time	$t_{RCH}$	0		0		0		ns	⑬
Write Command Hold Time	$t_{WCH}$	55		45		35		ns	
Write Command Hold Time Referenced to RAS	$t_{WCR}$	120		95		95		ns	
Write Command Pulse Width	$t_{WP}$	55		45		35		ns	
Write Command to RAS Lead Time	$t_{RWL}$	55		45		40		ns	
Write Command to CAS Lead Time	$t_{CWL}$	55		45		40		ns	
Data-In Set-Up Time	$t_{DS}$	0		0		0		ns	⑪
Data-In Hold Time	$t_{DH}$	55		45		35		ns	⑪
Data-In Hold Time Referenced to RAS	$t_{DHR}$	155		120		95		ns	
Refresh Period	$t_{REF}$		2		2		2	ms	
WRITE Command Set-Up Time	$t_{WCS}$	10		-10		-10		ns	⑭
CAS to WRITE Delay	$t_{CWD}$	55		45		40		ns	⑫
RAS to WRITE Delay	$t_{RWD}$	130		120		100		ns	⑫

- Notes: ①  $T_a$  is specified here for operation at frequencies to  $t_{RC} > t_{RC}(\text{min})$ . Operation at higher cycle rates with reduced ambient temperatures and higher power dissipation is permissible, however, provided AC operating parameters are met.
- ② An initial pause of 100  $\mu\text{s}$  is required after power-up followed by any 8 RAS cycles before proper device operation is achieved.
- ③ AC measurements assume  $t_T = 5\text{ ns}$ .
- ④  $V_{IH}(\text{min})$  or  $V_{IH}(\text{min})$  and  $V_{IL}(\text{max})$  are reference levels for measuring timing of input signals. Also, transition times are measured between  $V_{IHC}$  or  $V_{IHL}$  and  $V_{IL}$ .
- ⑤ The specifications for  $t_{RC}(\text{min})$  and  $t_{RWC}(\text{min})$  are used only to indicate cycle times at which proper operation over the full temperature range ( $0^\circ\text{C} < T_a < 70^\circ\text{C}$ ) is assured.
- ⑥ Assumes that  $t_{RCS} \leq t_{RCD}(\text{max})$ . If  $t_{RCS}$  is greater than the maximum recommended value shown in this table,  $t_{RAC}$  will increase by the amount that  $t_{RCD}$  exceeds the values shown.
- ⑦ Assumes that  $t_{RCD} > t_{RCD}(\text{max})$ .
- ⑧ Measured with a load equivalent to 2 TTL loads and 100 pF.
- ⑨  $t_{OFF}(\text{max})$  defines the time at which the output achieves the open circuit condition and is not referenced to output voltage levels.
- ⑩ Operation within the  $t_{RCD}(\text{max})$  limit ensures that  $t_{RAC}(\text{max})$  can be met,  $t_{RCD}(\text{max})$  is specified as a reference point only, if  $t_{RCD}$  is greater than the specified  $t_{RCD}(\text{max})$  limit, then access time is controlled exclusively by  $t_{CAC}$ .
- ⑪ These parameters are referenced to CAS leading edge in early write cycles and to WRITE leading edge in delayed write or read-modify-write cycles.
- ⑫  $t_{WCS}$ ,  $t_{CWD}$  and  $t_{RWD}$  are restrictive operating parameters in read-write and read-modify-write cycles only. If  $t_{WCS} > t_{WCS}(\text{min})$ , the cycle is an early write cycle and the data output will remain open circuit throughout the entire cycle. If  $t_{CWD} > t_{CWD}(\text{min})$  and  $t_{RWD} > t_{RWD}(\text{min})$ , the cycle is a read-write and the data output will contain data read from the selected cell. If neither of the above conditions are met the condition of the data out (set access time and until CAS goes back to  $V_{IH}$ ) is indeterminate.
- ⑬ Either  $t_{RRH}$  or  $t_{RCH}$  must be satisfied for a read cycle.

**Description**

The  $\mu$ PD2147A is a 4096-bit static Random Access Memory organized as 4096 words by 1 bit, using a scaled MOS technology. It uses a uniquely innovative design approach which provides the ease-of-use features associated with non-clocked static memories and the reduced standby power dissipation associated with clocked static memories. To the user this means low standby power dissipation without the need for clocks, address setup and hold times, nor reduced data rates due to cycle times that are longer than access times.

$\overline{CS}$  controls the power-down feature. In less than a cycle time after  $\overline{CS}$  goes high — deselection the  $\mu$ PD2147A — the part automatically reduces its power requirements and remains in this low power standby mode as long as  $\overline{CS}$  remains high.

This device feature results in system power savings as great as 85% in larger systems, where the majority of devices are deselected.

The  $\mu$ PD2147A is placed in an 18-pin cerdip package configured with the industry standard 2147 pinout. It is directly TTL compatible in all respects: input, output, and a single +5V supply. The data is read out nondestructively and has the same polarity as the input data. A data input and a separate three-state output are used.

**Features**

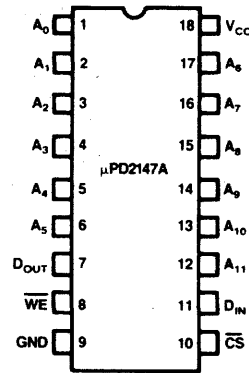
- Pinout, Function, and Power Compatible to Industry Standard 2147
- Scaled MOS Technology
- Completely Static Memory — No Clock or Timing Strobe Required
- Equal Access and Cycle Times
- Single +5V Supply
- Direct Performance Upgrade for 2147
- Automatic Power-Down
- High Density 18-Pin Package
- Directly TTL Compatible — All Input and Output
- Separate Data Input and Output
- Three-State Output

	Max Access Time	Supply Current	
		Active	Standby
$\mu$ PD2147A-25	25 ns	160 mA	20 mA
$\mu$ PD2147A-35	35 ns	160 mA	20 mA
$\mu$ PD2147A-45	45 ns	160 mA	20 mA

**Truth Table**

$\overline{CS}$	$\overline{WE}$	Mode	Output	Power
H	X	Not Selected	High Z	Standby
L	L	Write	High Z	Active
L	H	Read	$D_{OUT}$	Active

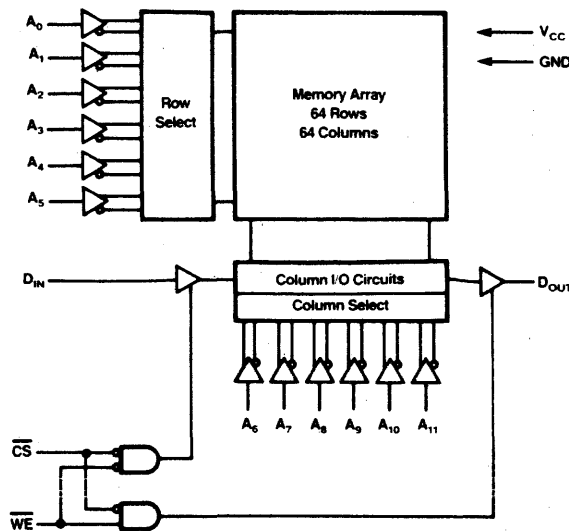
**Pin Configuration**



**Pin Identification**

Pin			Function
No.	Symbol		
1-6, 17-12	$A_0$ - $A_{11}$		Address Inputs
7	$D_{OUT}$		Data Output
8	$\overline{WE}$		Write Enable
9	GND		Ground
10	$\overline{CS}$		Chip Select
11	$D_{IN}$		Data Input
18	$V_{CC}$		Power (+5V)

**Block Diagram**



**Absolute Maximum Ratings\***

<b>T<sub>a</sub> = 25°C</b>	
Operating Temperature	-10°C to +85°C
Storage Temperature	-65°C to +150°C
Voltage on Any Pin	-3.5V to +7V ①
DC Output Current	20mA
Power Dissipation	1.2 W

Note: ① with respect to ground

\*COMMENT: Exposing the device to stresses above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational sections of this specification. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**DC and Operating Characteristics**

T<sub>a</sub> = 0°C to +70°C; V<sub>CC</sub> = +5V ± 10% ①

Parameter	Symbol	Limits			Unit	Test Conditions
		Min	Typ <sup>2</sup>	Max		
Input Load Current (All Input Pins)	I <sub>LI</sub>	0.01	10		μA	V <sub>CC</sub> = Max, V <sub>IN</sub> = GND to V <sub>CC</sub>
Output Leakage Current	I <sub>LO</sub>	0.01	10		μA	CS = V <sub>IH</sub> , V <sub>CC</sub> = Max, V <sub>OUT</sub> = GND to V <sub>CC</sub>
Operating Current	I <sub>CC</sub>	120	150	160	mA	T <sub>a</sub> = 25°C, V <sub>CC</sub> = Max, CS = V <sub>IL</sub> , Output Open
Standby Current	I <sub>SB</sub>	12	20		mA	V <sub>CC</sub> = Min to Max, CS = V <sub>IH</sub>
Peak Power-On Current	I <sub>PO</sub> <sup>3</sup>	25	50		mA	V <sub>CC</sub> = GND to V <sub>CC</sub> = Min, CS = Lower to V <sub>CC</sub> or V <sub>IH</sub> Min
Input Low Voltage	V <sub>IL</sub>	-3.0		0.8	V	
Input High Voltage	V <sub>IH</sub>		2.0	6.0	V	
Output Low Voltage	V <sub>OL</sub>			0.4	V	I <sub>OL</sub> = 8 mA
Output High Voltage	V <sub>OH</sub>		2.4		V	I <sub>OH</sub> = -4.0 mA
Output Short Circuit Current	I <sub>OS</sub>			±130	mA	V <sub>OUT</sub> = GND to V <sub>CC</sub>

- Notes: ① The operating ambient temperature range is guaranteed with transverse air flow exceeding 400 linear feet per minute.  
 ② Typical limits are V<sub>CC</sub> = 5V, T<sub>a</sub> = 25°C, and specified loading.  
 ③ A pull-up resistor to V<sub>CC</sub> on the CS input is required to keep the device deselected; otherwise, power-on current approaches I<sub>CC</sub> active.

**AC Test Conditions**

Input Pulse Levels	GND to 3.0V
Input Rise and Fall Times	5ns
Input Timing Reference Levels	1.5V
Output Timing Reference Levels	1.5V
Output Load	See Figure 1

**Capacitance**

T<sub>a</sub> = 25°C; f = 1.0 MHz ①

Parameter	Symbol	Limits			Unit	Test Conditions
		Min	Typ	Max		
Input Capacitance	C <sub>IN</sub>		5		pF	V <sub>IN</sub> = 0V
Output Capacitance	C <sub>OUT</sub>		6		pF	V <sub>OUT</sub> = 0V

Note: ① This parameter is sampled and not 100% tested.

Figure 1 — Output Load

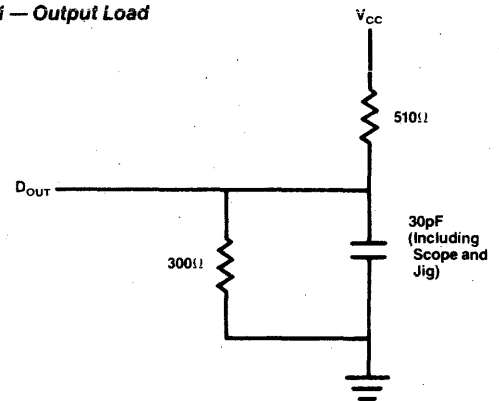
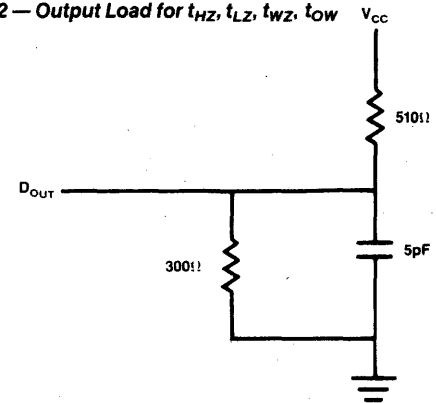


Figure 2 — Output Load for t<sub>HZ</sub>, t<sub>LZ</sub>, t<sub>WZ</sub>, t<sub>OW</sub>



**AC Characteristics**

**Read Cycle**

$T_a = 0^\circ\text{C to } +70^\circ\text{C}; V_{CC} = +5V \pm 10\%$ , unless otherwise noted.

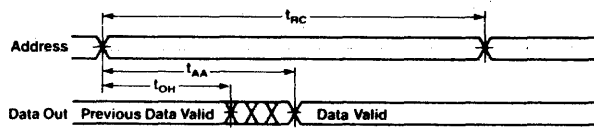
Parameter	Symbol	Limits						Unit	Test Conditions
		μPD2147A-25		μPD2147A-35		μPD2147A-45			
		Min	Max	Min	Max	Min	Max		
Read Cycle Time	$t_{RC}^1$	25	35	35	45	45	ns		
Address Access Time	$t_{AA}$	25	35	35	45	45	ns		
Chip Select Access Time	$t_{ACS}$	25	35	35	45	45	ns		
Output Hold From Address Change	$t_{OH}$	5	5	5	5	5	ns		
Chip Select to Output in Low Z	$t_{LZ}^2$	5	5	5	5	5	ns	③	
Chip Deselection to Output in High Z	$t_{HZ}^2$	0	20	0	30	0	30	ns	④
Chip Selection to Power-Up Time	$t_{PU}$	0	0	0	0	0	ns		
Chip Selection to Power-Down Time	$t_{PD}$	20	20	20	20	20	ns		

**Write Cycle**

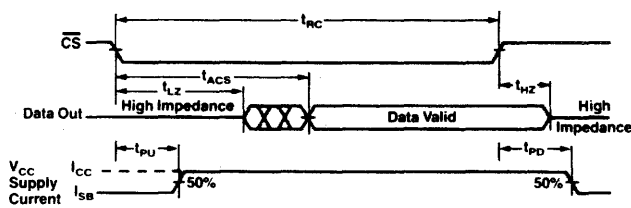
Parameter	Symbol	Limits						Unit	Test Conditions
		μPD2147A-25		μPD2147A-35		μPD2147A-45			
		Min	Max	Min	Max	Min	Max		
Write Cycle Time <sup>②</sup>	$t_{WC}$	25	35	35	45	45	ns		
Chip Selection to End of Write	$t_{CW}$	25	35	35	45	45	ns		
Address Valid to End of Write	$t_{AW}$	25	35	35	45	45	ns		
Address Setup Time	$t_{AS}$	0	0	0	0	0	ns		
Write Pulse Width	$t_{WP}$	20	20	20	25	25	ns		
Write Recovery Time	$t_{WR}$	0	0	0	0	0	ns		
Data Valid to End of Write	$t_{DW}$	20	20	20	25	25	ns		
Data Hold Time	$t_{DH}$	10	10	10	10	10	ns		
Write Enabled to Output in High Z	$t_{WZ}$	0	15	0	20	0	25	ns	③
Output Active From End of Write	$t_{OW}$	0	0	0	0	0	ns	④	

**Timing Waveforms**

**Read Cycle No. 1** ⑤⑥



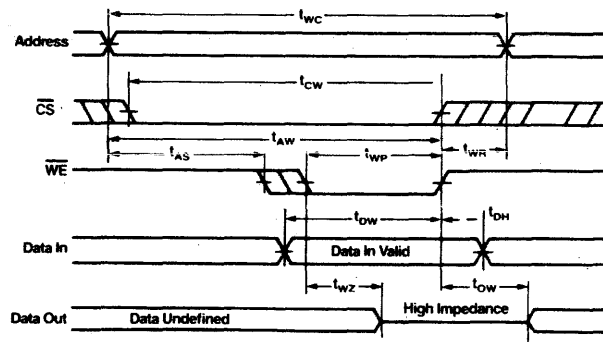
**Read Cycle No. 2** ⑤⑦



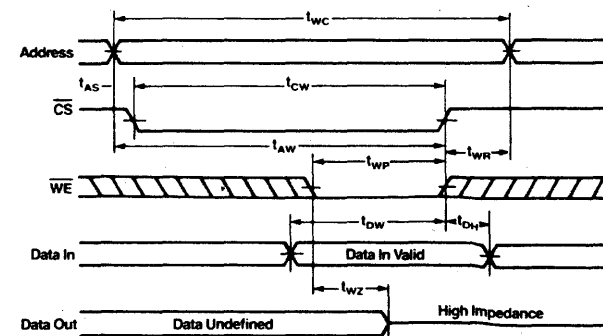
**Notes:**

- All Read Cycle timings are referenced from the last valid address to the first transitioning address.
- At any given temperature and voltage condition,  $t_{HZ}$  max is less than  $t_{LZ}$  min, both for a given device and from device to device.
- Transition is measured - 200mV from steady state voltage with specified loading in Figure 2.
- Transition is measured at  $V_{OL} = 200mV$  and  $V_{OH} = 200mV$  with specified loading in Figure 2.
- WE is high for Read Cycles.
- Device is continuously selected.  $\overline{CS} = V_{IL}$ .
- Addresses valid prior to or coincident with  $\overline{CS}$  transition low.

**Write Cycle No. 1** ( $\overline{WE}$  Controlled) ⑤



**Write Cycle No. 2** ( $\overline{CS}$  Controlled) ⑤



**Notes:**

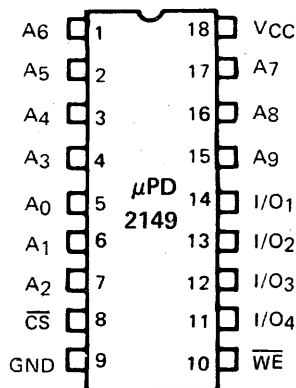
- If  $\overline{CS}$  goes high simultaneously with  $\overline{WE}$  high, the output remains in a high impedance state.
- All Write Cycle timings are referenced from the last valid address to the first transitioning address.
- Transition is measured at  $V_{OL} = 200mV$  and  $V_{OH} = 200mV$  with specified loading in Figure 2.
- Transition is measured - 200mV from steady state voltage with specified loading in Figure 2.
- $\overline{CS}$  or  $\overline{WE}$  must be high during address transitions.

**4096 (1024x4) BIT STATIC RAM**

**DESCRIPTION** The  $\mu$ PD2149 is a 4096-bit static Random Access Memory organized as 1024 words by 4-bits. Using a scaled NMOS technology, it incorporates an innovative design approach which provides the ease-of-use features associated with non-clocked static memories.

The  $\mu$ PD2149 is encapsulated in an 18-pin ceramic package configured with the industry standard pinout. It is directly TTL compatible in all respects: inputs, outputs, and a single +5V supply. The data is read out non-destructively and has the same polarity as the input data.

- FEATURES**
- Completely Static Memory – No Clock or Timing Strobe Required
  - Equal Access and Cycle Times, Faster Chip Select Access
  - Single +5V Supply
  - High Density 18-Pin Package
  - Directly TTL Compatible – All Inputs and Outputs
  - Common Input and Output
  - Three-State Output
  - Access Time: 35-55 ns MAX (From Address)  
15-25 ns MAX (From Chip Select)
  - Power Dissipation: 180 mA MAX

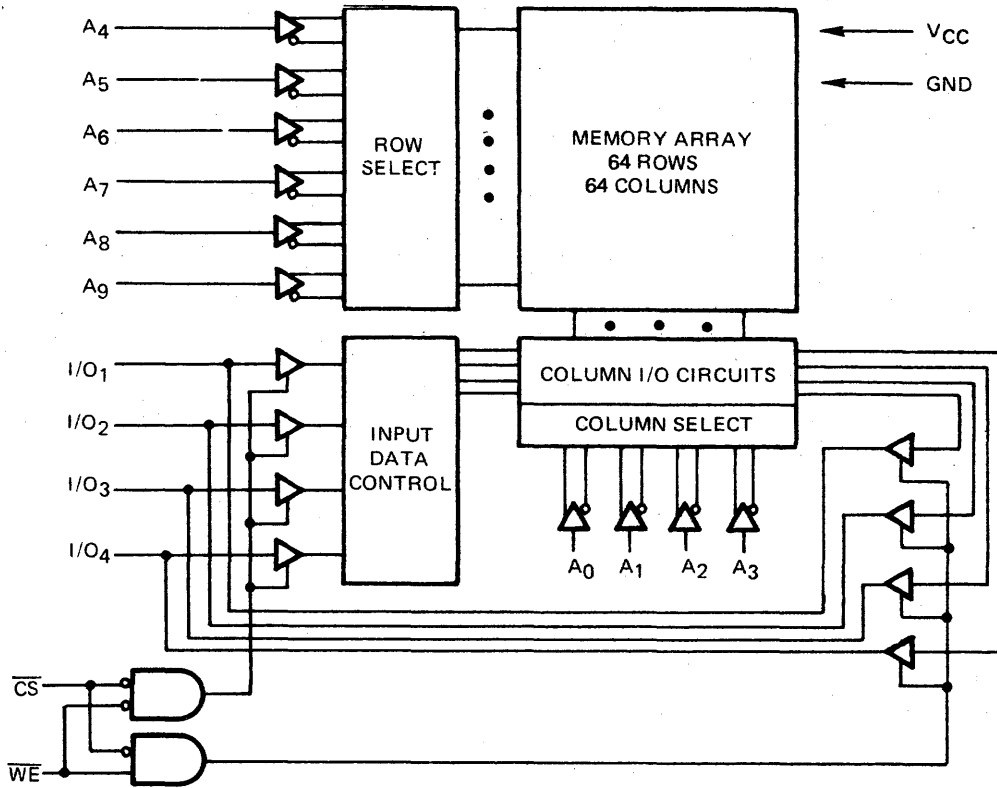


**PIN NAMES**

A <sub>0</sub> -A <sub>9</sub>	Address Inputs
$\overline{WE}$	Write Enable
$\overline{CS}$	Chip Select
I/O <sub>1</sub> -I/O <sub>4</sub>	Data Input/Output
V <sub>CC</sub>	Power (+5V)
GND	Ground

**TRUTH TABLE**

$\overline{CS}$	$\overline{WE}$	MODE	I/O
H	X	Not Selected	High Z
L	L	Write	D <sub>IN</sub>
L	H	Read	D <sub>OUT</sub>



- Operating Temperature . . . . . -10°C to +85°C
- Storage Temperature . . . . . -65°C to +150°C
- Voltage on Any Pin . . . . . -1.5V to +7V ①
- DC Output Current . . . . . 20 mA
- Power Dissipation . . . . . 1.2W

**ABSOLUTE MAXIMUM RATINGS\***

Note: ① with respect to ground

T<sub>a</sub> = 25°C

\*COMMENT: Stress above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

T<sub>a</sub> = 0°C to +70°C; V<sub>CC</sub> = +5V ± 10%, unless otherwise noted.

**DC CHARACTERISTICS**

PARAMETER	SYMBOL	MIN	MAX	UNIT	TEST CONDITIONS
Input Leakage Current	I <sub>LI</sub>		+10	μA	V <sub>IN</sub> = GND to V <sub>CC</sub>
Output Leakage Current	I <sub>LO</sub>		+50	μA	CS = V <sub>IH</sub> V <sub>OUT</sub> = GND to 4.5V
Power Supply Current	I <sub>CC</sub>		180	MA	V <sub>IN</sub> = V <sub>CC</sub> , I/O = open
Input Low Voltage	V <sub>IL</sub>		0.8	V	
Input High Voltage	V <sub>IH</sub>	2.1	V <sub>CC</sub>	V	
Output Low Voltage	V <sub>OL</sub>		0.4	V	I <sub>OL</sub> = 8 MA
Output High Voltage	V <sub>OH</sub>	2.4		V	I <sub>OH</sub> = -4 MA
Output Short Circuit Current	I <sub>OS</sub>		±200	MA	V <sub>OUT</sub> = GND to V <sub>CC</sub>

Note: The operating temperature range is guaranteed with transverse air flow exceeding 400 feet per minute.

MEMORY  
NEC Electronics



**CAPACITANCE**  $T_a = 25^\circ\text{C}; f = 1.0\text{ MHz}$  ①

PARAMETER	SYMBOL	LIMITS			UNIT	TEST CONDITIONS
		MIN	TYP	MAX		
Input Capacitance	$C_{IN}$			5	pF	$V_{IN} = 0V$
Output Capacitance	$C_{OUT}$			7	pF	$V_{OUT} = 0V$

Note: ① This parameter is sampled and not 100% tested.

**AC TEST CONDITIONS**

Input Pulse Levels ..... Gnd to 3.0V  
 Input Rise and Fall Times ..... 5 ns  
 Input and Output Timing Reference Levels ..... 1.5V  
 Output Load ..... See Figure 1

**AC CHARACTERISTICS READ CYCLE** ①

$T_a = 0^\circ\text{C}$  to  $+70^\circ\text{C}$ ;  $V_{CC} = +5V \pm 10\%$ , unless otherwise noted.

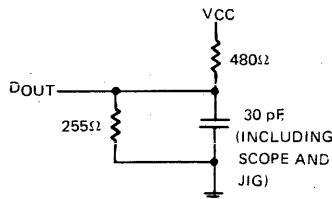


Figure 1

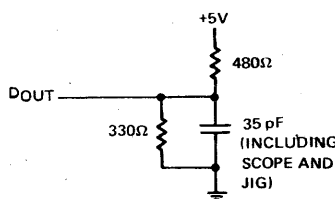


Figure 2

PARAMETER	SYMBOL	2149-2		2149-1		2149		UNIT	TEST CONDITIONS
		MIN	MAX	MIN	MAX	MIN	MAX		
Read Cycle Time	$T_{RC}$	35		45		55		ns	
Access Time	$T_A$		35		45		55	ns	
Chip Selection to Output Valid	$T_{CO}$		15		20		25	ns	
Chip Selection to Output Active	$T_{CX}$	5		5		5		ns	
Output 3-State From Deselection	$T_{OTD}$		10		15		20	ns	②
Output Hold From Address Change	$T_{OH}$	5		5		5		ns	

Notes: ①  $\overline{WE}$  is high for read cycle.

② Transition is measured +500 MV from steady state with load of Figure 2. This parameter is sampled and not 100% tested.

PARAMETER	SYMBOL	μPD2149-2		μPD2149-1		μPD2149		UNIT	TEST CONDITIONS
		MIN	MAX	MIN	MAX	MIN	MAX		
Write Cycle Time	$t_{WC}$	35		45		55		ns	
Chip Selection to End of Write	$t_{CW}$	30		40		50		ns	
Address Valid to End of Write	$t_{AW}$	30		40		50		ns	
Address Setup Time	$t_{AS}$	0		0		0		ns	
Write Pulse Width	$t_{WP}$	30		35		40		ns	
Write Recovery Time	$t_{WR}$	5		5		5		ns	
Data Valid to End of Write	$t_{DW}$	20		20		20		ns	
Data Hold Time	$t_{DH}$	5		5		5		ns	
Write Enabled to Output in High Z	$t_{WZ}$	0	10	0	15	0	20	ns	①
Output Active from End of Write	$t_{OW}$	0		0		0		ns	②

Notes: ①  $WE$  or  $CS$  must be high during all address transitions.

② Transition is measured +500 MV from steady state with load of Figure 3. This parameter is sampled and not 100% tested.

**AC CHARACTERISTICS WRITE CYCLE**

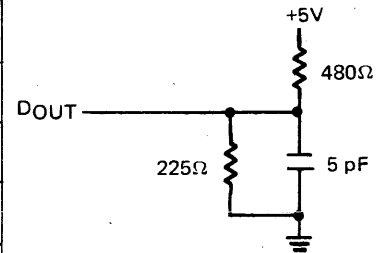


Figure 3

**Description**

The μPD2167 is a 16,384-word by 1-bit static MOS RAM. Using a scaled-NMOS technology, its design provides the easy-to-use features associated with non-clocked static memories.

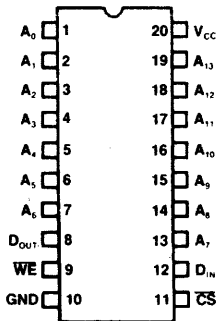
The μPD2167 has a three-state output and offers a standby mode that features an 83% savings in power consumption. The μPD2167 requires a single +5 volt supply and is fully TTL-compatible. It features equal access and cycle times and, because of its fully static operation, it requires no external clocks or timing strobes. It is packaged in a standard 20-pin, 300 mil DIP.

**Features**

- 16384 x 1 organization
- Fully static memory — no clock or timing strobe required
- Equal access and cycle times
- Single +5v supply
- Automatic power-down
- Standard 20-pin DIP, 300 mil
- All inputs and output directly TTL-compatible
- Separate data input and output
- Three-state output
- Power dissipation: 180 mA max (active)  
30 mA max (standby)

	Access time	R/W Cycle time
μPD2167-3	55ns	55ns
μPD2167-2	70ns	70ns

**Pin Configuration**

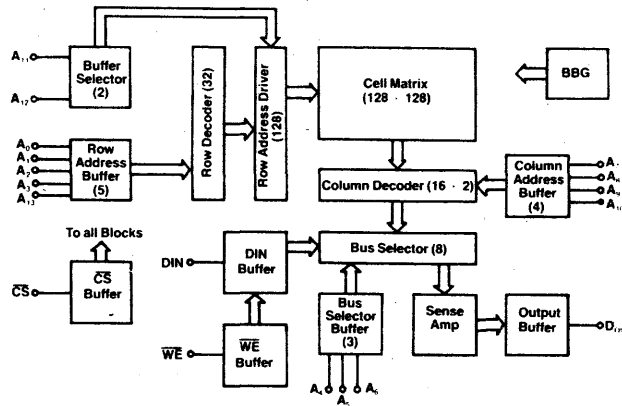


**Pin Names**

A <sub>0</sub> -A <sub>13</sub>	Address Inputs
WE	Write Enable
CS	Chip Select
D <sub>IN</sub>	Data Input
D <sub>OUT</sub>	Data Output
V <sub>CC</sub>	Power (+5v)
V <sub>SS</sub>	Ground

**Truth Table**

CS	WE	Mode	Output	Power
H	X	not selected	High Z	Standby
L	L	write	High Z	Active
L	H	read	D <sub>OUT</sub>	Active



**Absolute Maximum Ratings\***

T <sub>a</sub> = 25°C	
Temperature under bias	-10°C to +85°C
Storage temperature	-65°C to +150°C
Voltage on any pin with respect to ground	-3.5v to +7v
D.C. output current	20mA
Power dissipation	1.2w

\* Comment: Exposing the device to stresses above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational sections of this specification. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Capacitance**

T<sub>a</sub> = 25°C, f = 1.0 MHz

Parameter	Symbol	Max.	Unit	Conditions
Input Capacitance	C <sub>IN</sub>	5	pF	V <sub>IN</sub> = 0V
Output Capacitance	C <sub>OUT</sub>	6	pF	V <sub>OUT</sub> = 0V

This parameter is sampled and not 100% tested.

**DC Characteristics**

$T_a = 0^\circ\text{C to } +70^\circ\text{C}, V_{CC} = +5\text{v} \pm 10\%$

Parameter	Sym	Min	Typ	Max	Unit	Test Conditions
Input load current all input pins	$I_{LI}$			10	$\mu\text{A}$	$V_{CC} = \text{max}, V_{IN} = \text{GND to } V_{CC}$
Output leakage current	$I_{LO}$		0.1	50	$\mu\text{A}$	$\overline{\text{CS}} = V_{IH}, V_{CC} = \text{max}, V_{OUT} = \text{GND to } V_{CC}$
Operating current	$I_{CC}$			170	$\text{mA}$	$T_A = 25^\circ\text{C}, V_{CC} = \text{max}, \overline{\text{CS}} = V_{IL}, \text{output open}$
				180	$\text{mA}$	$T_A = 0^\circ\text{C}$
Standby current	$I_{SB}$			30	$\text{mA}$	$V_{CC} = \text{min to max}, \overline{\text{CS}} = V_{IH}$
Peak Power-On current	$I_{PO} \textcircled{1}$		35	70	$\text{mA}$	$V_{CC} = \text{GND to } V_{CC} \text{ min. } \overline{\text{CS}} = \text{Lower of } V_{CC} \text{ or } V_{IH} \text{ min.}$
Input low voltage	$V_{IL}$	-3.0		0.8	$\text{V}$	
Input high voltage	$V_{IH}$	2.0		6.0	$\text{V}$	
Output low voltage	$V_{OL}$			0.4	$\text{V}$	$I_{OL} = 8 \text{ mA}$
Output high voltage	$V_{OH}$	2.4			$\text{V}$	$I_{OH} = -4 \text{ mA}$
Output short circuit current	$I_{OS1}$		-150		$\text{mA}$	$V_{OUT} = \text{GND}$
Output short circuit current	$I_{OS2}$		150		$\text{mA}$	$V_{OUT} = V_{CC}$

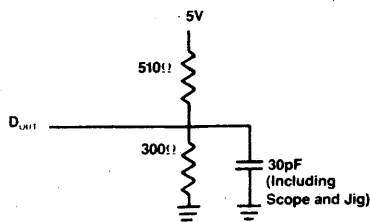


Figure 1 - Output Load

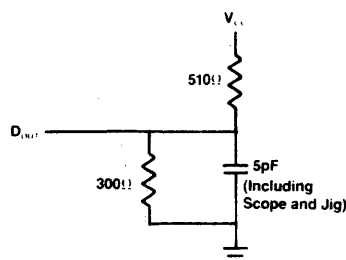


Figure 2 - Output Load for  $t_{HZ}, t_{LZ}, t_{WZ}, t_{OW}$

**AC Characteristics**

$T_a = 0^\circ\text{C to } +70^\circ\text{C}, V_{CC} = +5\text{V} \pm 10\%$

Parameter	Symbol	μPD2167-3		μPD2167-2		Unit	Notes
		min	typ	min	typ		
<b>Read Cycle</b>							
Read cycle time	$t_{RC}$	55		70		ns	
Address access time	$t_{AA}$		55		70	ns	①
Chip select access time	$t_{ACS}$		55		70	ns	②
Output hold from address change	$t_{OH}$	5		5		ns	
Chip select to output in low Z	$t_{LZ}$	10		10		ns	
Chip deselect to output in high Z	$t_{HZ}$	0	40	0	40	ns	
Chip select to power up time	$t_{PU}$	0		0		ns	
Chip deselect to power down time	$t_{PD}$	30		30		ns	
<b>Write Cycle</b>							
Write cycle time	$t_{WC}$	55		70		ns	
Chip select to end of write	$t_{CW}$	45		55		ns	
Address valid to end of write	$t_{AW}$	45		55		ns	
Address setup time	$t_{AS}$	0		0		ns	
Write pulse	$t_{WP}$	35		40		ns	
Write recovery time	$t_{WR}$	10		15		ns	
Data valid to end of write	$t_{DW}$	25		30		ns	
Data hold time	$t_{DH}$	10		10		ns	
Write enabled to output in high Z	$t_{WZ}$	0	30	0	35	ns	
Output active from end of write	$t_{OW}$	0		0		ns	

**Notes:**

- ① CS valid prior to or coincident with address transition
- ② Address valid prior to or coincident with CS transition low

**Description**

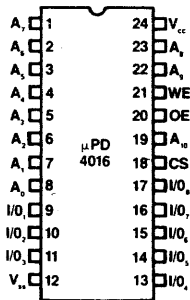
The μPD4016 is a 16384-bit static Random Access Memory device organized as 2048 words by 8 bits. Using a scaled NMOS technology, its design provides the ease-of-use features associated with nonclocked static memories. The μPD4016 has a three-state output and offers a standby mode with an attendant 75% to 90% savings in power consumption. It features equal access and cycle times and provides an output enable function that eliminates the need for external bus buffers. The μPD4016 is packaged in 600 mil and 300 mil 24-pin dual-in-line packages and is plug-compatible with 2716 EPROMS.

**Features**

- Scaled NMOS technology
- Completely static memory: no clock, no refresh
- Equal access and cycle times
- Single +5V supply
- Automatic power-down
- All inputs and outputs directly TTL-compatible
- Common I/O capability
- OE eliminates need for external bus buffers
- Three-state outputs
- Plug-compatible with 2716 EPROMS
- Standby current 15mA standard, selectable to 5mA
- Available in 600 mil and 300 mil DIP (Skinnydip)

	Access Time	R/W Cycle	I <sub>cc</sub> (max)
μPD4016-1	250 ns	250 ns	60 mA
μPD4016-2	200 ns	200 ns	60 mA
μPD4016-3	150 ns	150 ns	60 mA

**Pin Configuration**



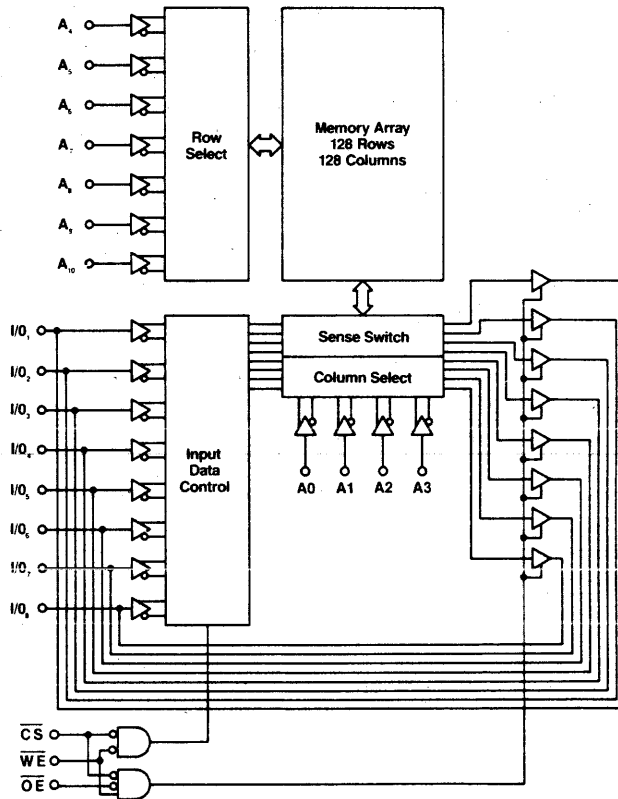
**Pin Identification**

No.	Symbol	Function
1-8, 19, 22, 23	A <sub>7</sub> -A <sub>0</sub> , A <sub>10</sub> , A <sub>9</sub> , A <sub>8</sub>	Address Inputs
9-11, 13-17	I/O <sub>1</sub> - I/O <sub>8</sub>	Data Input/Output
12	V <sub>SS</sub>	Ground
18	CS	Chip Select
20	OE	Output Enable
21	WE	Write Enable
24	V <sub>CC</sub>	Power (+5V)

**Truth Table**

CS	OE	WE	Mode	I/O	Power
H	X	X	Not Selected	High-Z	Standby
L	L	H	Read	D <sub>OUT</sub>	Active
L	H	L	Write	D <sub>IN</sub>	Active
L	L	L	Write	D <sub>IN</sub>	Active

**Block Diagram**



**Absolute Maximum Ratings\***

<b>T<sub>a</sub> = 25°C</b>	
Temperature Under Bias	-10°C to +85°C
Storage Temperature	-65°C to +150°C (4016D) -55°C to +125°C (4016C)
Voltage on Any Pin with Respect to Ground	-0.5V to +7V
DC Output Current	20mA
Power Dissipation	1W

\*COMMENT: Exposing the device to stresses above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational sections of this specification. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Rev/3

**Capacitance**  
 $T_a = 25^\circ\text{C}; f = 1\text{ MHz}$

Parameter	Symbol	Limits			Unit	Test Conditions
		Min	Typ	Max		
Input Capacitance	$C_{in}$			5	pF	$V_{in} = 0V$
I/O Capacitance	$C_{fb}$			7	pF	$V_{fb} = 0V$

This parameter is sampled and not 100% tested.

**DC Characteristics**  
 $T_a = 0^\circ\text{C to } 70^\circ\text{C}, V_{CC} = 5V \pm 10\%$

Parameter	Symbol	Limits			Unit	Test Conditions
		Min	Typ	Max		
Input Leakage Current	$I_{LI}$			10	$\mu\text{A}$	$V_{CC} = \text{Max.}$ $V_{IN} = \text{GND to } V_{CC}$
Output Leakage Current	$I_{LO}$			10	$\mu\text{A}$	$V_{CC} = \text{Max. CS} = V_{IH}$ $V_{OUT} = \text{GND to } V_{CC}$
Operating Current	$I_{CC}$			60	mA	$V_{CC} = \text{Max. CS} = V_{IL}$ Outputs Open
Standby Current ①	$I_{SB}$			15	mA	$V_{CC} = \text{Min. to Max.}$ $CS = V_{IH}$
Input Low Voltage	$V_{IL}$	-0.3		0.8	V	
Input High Voltage	$V_{IH}$	2.0		6.0	V	
Output Low Voltage	$V_{OL}$			0.4	V	$I_{OL} = 4\text{mA}$
Output High Voltage	$V_{OH}$	2.4			V	$I_{OH} = 1\text{mA}$
Output Short Circuit Current	$I_{OS}$	TBD		TBD	mA	$V_{OUT} = \text{GND to } V_{CC}$

Notes: ① 5mA max available (contact regional office).

**AC Test Conditions**

Input Pulse Levels	0.8V to 2.2V
Input Rise and Fall Times	10ns
Input Timing Reference Levels	1.5V
Output Timing Reference Levels	1.5V

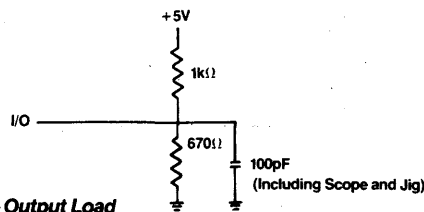


Figure 1 - Output Load

**AC Characteristics**

**Read Cycle**  
 $T_a = 0^\circ\text{C to } 70^\circ\text{C}, V_{CC} = 5V \pm 10\%$

Parameter	Symbol	Limits						Unit	Test Conditions
		μPD4016-3		μPD4016-2		μPD4016-1			
		Min	Max	Min	Max	Min	Max		
Read Cycle Time	$t_{RC}$	150		200		250		ns	①
Address Access Time	$t_{AA}$		150		200		250	ns	
Chip Select Access Time	$t_{ACS}$		150		200		250	ns	②
Output Hold from Address Change	$t_{OH}$	10		10		10		ns	
Chip Selection to Output in Low Z	$t_{LZ}$	10		10		10		ns	③④
Chip Deselection to Output in High Z	$t_{HZ}$		50		60		80	ns	③④
Output Enable to Output Valid	$t_{OE}$		70		90		110	ns	
Output Enable to Output in Low Z	$t_{OLZ}$	10		10		10		ns	③④
Output Disable to Output in High Z	$t_{OHZ}$		50		60		80	ns	③④
Chip Selection to Power up Time	$t_{PU}$	0		0		0		ns	⑤
Chip Deselection to Power down Time	$t_{PD}$		70		90		110	ns	⑥

**Write Cycle**

$T_a = 0^\circ\text{C to } 70^\circ\text{C}, V_{CC} = 5V \pm 10\%$

Parameter	Symbol	Limits						Unit	Test Conditions
		μPD4016-3		μPD4016-2		μPD4016-1			
		Min	Max	Min	Max	Min	Max		
Write Cycle Time	$t_{WC}$	150		200		250		ns	
Chip Selection to End of Write	$t_{CW}$	120		160		200		ns	
Address Valid to End of Write	$t_{AW}$	90		120		150		ns	
Address Setup Time	$t_{AS}$	0		0		0		ns	
Write Pulse Width	$t_{WP}$	80		100		130		ns	⑤
Write Recovery Time	$t_{WR}$	10		10		10		ns	
Data Valid to End of Write	$t_{OW}$	50		60		80		ns	
Data Hold Time	$t_{OH}$	0		0		0		ns	
Write Enabled to Output in High-Z	$t_{WZ}$		50		60		80	ns	⑥⑦
Output Active from End of Write	$t_{OW}$	10		10		10		ns	⑥⑦

**Notes:**

- ① All Read Cycle timings are referenced from the last valid address to the first transitioning address.
- ② Address valid prior to or coincident with CS transition low.
- ③ Transition is measured  $\pm 200\text{mV}$  from steady state voltage with specified load of Figure 1.
- ④ This parameter is sampled and not 100% tested.
- ⑤ If CS and OE are both low before write enabled,  $t_{WP} = t_{WZ} + t_{OW}$ .
- ⑥ Transition is measured  $\pm 200\text{mV}$  from steady state voltage with specified loading in Figure 2.
- ⑦ This parameter is sampled and not 100% tested.

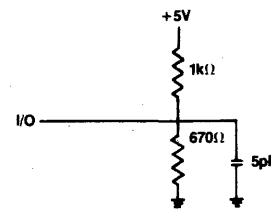


Figure 2 - Transition Load

**DESCRIPTION**

The μPD446 is a high speed, low power, 2048 word by 8 bit static CMOS RAM fabricated using an advanced silicon gate CMOS technology. A unique circuitry technique makes the μPD446 a very low operating power device which requires no clock or refreshing to operate. Minimum standby power current is drawn by this device when  $\overline{CS}$  equals  $V_{CC}$  independently of the other input levels.

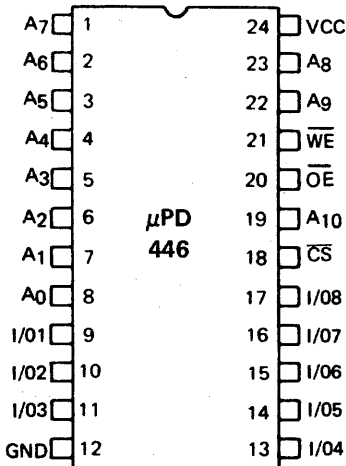
Data retention is guaranteed at a power supply voltage as low as 2V.

The μPD446 is packaged in a standard 24-pin dual-in-line package and is plug-in compatible with 16K EPROMs. The μPD446 is also available in the 24-pin mini-flat (SOP) package.

**FEATURES**

- Single +5V Supply
- Fully Static Operation – No Clock or Refreshing required
- TTL Compatible – All Inputs and Outputs
- Common I/O Using Three-State Output
- $\overline{OE}$  Eliminates Need for External Bus Buffers
- Max Access/Min Cycle Times Down to 150 ns
- Low power Dissipation, 18 mA Max Active/10 μA Max Standby/10 μA Max Data Retention
- Data Retention Voltage – 2V Min
- Standard 24-Pin Plastic, DIP, and Mini-flat (SOP) Packages
- Plug-in Compatible with 16K EPROMs
- Operating Temperature Range – -40°C to +85°C
- For operation at less than 4.5V power supply, contact the NEC Sales Office

**PIN CONFIGURATION**



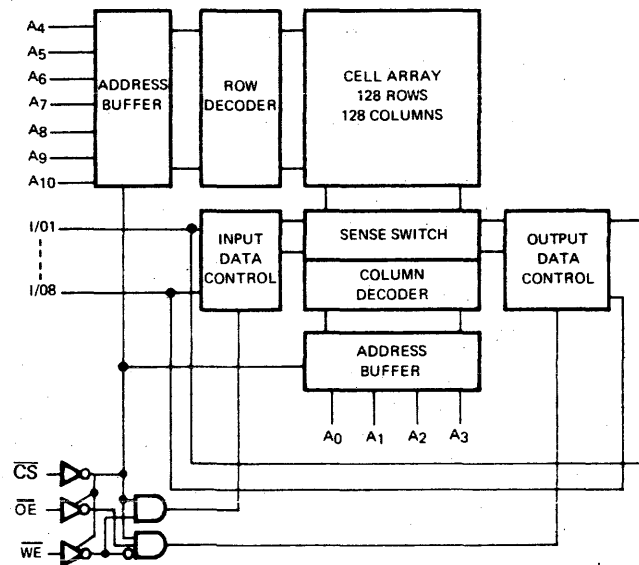
**PIN NAMES**

A <sub>0</sub> -A <sub>10</sub>	Address Inputs
$\overline{WE}$	Write Enable
$\overline{OE}$	Output Enable
$\overline{CS}$	Chip Select
I/O1-I/O8	Data Input/Output
VCC	Power (+5V)
GND	Ground

**TRUTH TABLE**

$\overline{CS}$	$\overline{OE}$	$\overline{WE}$	MODE	I/O	ICC
H	X	X	NOT SELECTED	HZ	STANDBY
L	H	H	NOT SELECTED	HZ	ACTIVE
L	L	H	READ	D <sub>OUT</sub>	ACTIVE
L	X	L	WRITE	D <sub>IN</sub>	ACTIVE

Rev/1



Supply Voltage ..... 7.0V  
 Input or Output Voltage Supplied ..... -0.3 to  $V_{CC} + 0.3V$   
 Storage Temperature Range ..... -55°C to 125°C  
 Operating Temperature Range ..... -40°C to +85°C

**ABSOLUTE MAXIMUM RATINGS\***

$T_a = 25^\circ C$

\*COMMENT: Stress above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

$T_a = -40^\circ C$  to  $+85^\circ C$ ,  $V_{CC} = 5.0V \pm 10\%$

DC CHARACTERISTICS

PARAMETER	SYMBOL	LIMITS			UNIT	CONDITIONS
		MIN	TYP	MAX		
Input High Voltage	$V_{IH}$	2.2		$V_{CC} + 0.3$	V	
Input Low Voltage	$V_{IL}$	-0.3		0.8	V	
Input Leakage Current	$I_{LI}$	-1.0		1.0	$\mu A$	$V_{IN} = 0 \sim V_{CC}$
I/O Leakage Current	$I_{LO}$	-1.0		1.0	$\mu A$	$V_{CS} = V_{IH}$ $V_{I/O} = 0 \sim V_{CC}$
Operating Supply Current	$I_{CCA1}$		①	①	mA	$V_{CS} = V_{IL}$ $I_{I/O} = 0$ MIN TCYCLE
	$I_{CCA2}$		5	10	mA	$V_{CS} = V_{IL}$ $I_{I/O} = 0$ DC CURRENT
	$I_{CCA3}$		30	100	$\mu A$	$V_{CS} = 0.2V$ $I_{I/O} = 0$ $V_{IN} = V_{CC} - 0.2$ OR 0.2V
Standby Current	$I_{CCS}$			10	$\mu A$	$V_{CS} = V_{CC} - 0.2$ $V_{IN} = 0 \sim V_{CC}$
Output High Voltage	$V_{OH}$	2.4			V	$I_{OH} = -1.0$ mA
Output Low Voltage	$V_{OL}$			0.4	V	$I_{OL} = 2.0$ mA

NOTE: ① μPD446: 12 mA TYP, 18 mA MAX  
 μPD446-1: 18 mA TYP, 26 mA MAX  
 μPD446-2: 20 mA TYP, 30 mA MAX  
 μPD446-3: 25 mA TYP, 38 mA MAX

CAPACITANCE  $T_a = 25^\circ\text{C}, f = 1.0\text{ MHz}$

PARAMETER	SYMBOL	LIMITS		UNIT	TEST CONDITIONS
		MIN	MAX		
Input Capacitance	$C_{IN}$		6	pF	$V_{IN} = 0V$
Input/Output Capacitance	$C_{I/O}$		8	pF	$V_{I/O} = 0V$

AC CHARACTERISTICS

READ CYCLE

$V_{CC} = 5.0V \pm 10\%, T_a = -40^\circ\text{C} \text{ to } +85^\circ\text{C}$

PARAMETER	SYMBOL	LIMITS								UNIT
		μPD446-3		μPD446-2		μPD446-1		μPD446		
		MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Read Cycle Time	$t_{RC}$	150		200		250		450		ns
Address Access Time	$t_{AA}$		150		200		250		450	ns
Chip Select Access Time	$t_{ACS}$		150		200		250		450	ns
Output Enable to Output Valid	$t_{OE}$		75		100		120		150	ns
Output Hold from Address Change	$t_{OH}$	15		15		15		15		ns
Chip Enable to Output in LZ	$t_{CLZ}$	10		10		10		10		ns
Output Enable to Output in LZ	$t_{OLZ}$	5		5		5		5		ns
Chip Disable to Output in HZ	$t_{CHZ}$		50		60		80		100	ns
Output Disable to Output in HZ	$t_{OHZ}$		50		60		80		100	ns

WRITE CYCLE

$V_{CC} = 5.0V \pm 10\%, T_a = -40^\circ\text{C} \text{ to } +85^\circ\text{C}$

PARAMETER	SYMBOL	LIMITS								UNIT
		μPD446-3		μPD446-2		μPD446-1		μPD446		
		MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Write Cycle Time	$t_{WC}$	150		200		250		450		ns
Chip Enable to End of Write	$t_{CW}$	120		150		180		210		ns
Address Valid to End of Write	$t_{AW}$	120		150		180		210		ns
Address Setup Time	$t_{AS}$	0		0		0		0		ns
Write Pulsewidth	$t_{WP}$	90		120		150		180		ns
Write Recovery Time	$t_{WR}$	0		0		0		0		ns
Data Valid to End of Write	$t_{DW}$	50		60		80		100		ns
Data Hold Time	$t_{DH}$	0		0		0		0		ns
Write Enable to Output in HZ	$t_{WHZ}$		50		60		80		100	ns
Output Active from End of Write	$t_{OW}$	10		10		10		10		ns

LOW  $V_{CC}$  DATA RETENTION

$T_a = -40^\circ\text{C} \text{ to } +85^\circ\text{C}$

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
$V_{CC}$ for Data Retention	$V_{CCDR}$	$V_{IN} = 0 \sim V_{CC}, V_{CS} = V_{CC}$	2.0			V
Data Retention Current	$I_{CCDR}$	$V_{CC} = 3.0V, V_{IN} = 0 \sim V_{CC}, V_{CS} = V_{CC}$		0.01	10	μA
Chip Deselection to Data Retention Time	$t_{CDR}$		0			ns
Operation Recovery Time	$t_R$		$t_{RC}$			ns



**2048 x 8-BIT STATIC CMOS RAM**

**DESCRIPTION** The μPD449 is a high speed, low power, 2048 word by 8-bit static CMOS RAM fabricated using an advanced silicon gate CMOS technology. A unique circuitry technique makes the μPD449 a very low operating power device which requires no clock or refreshing to operate.

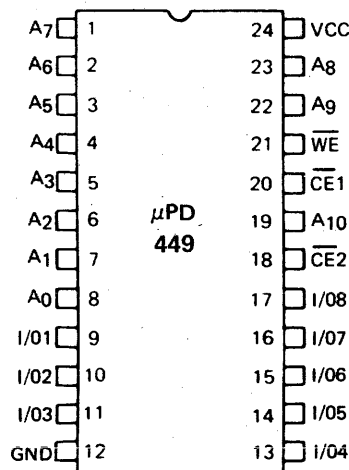
Since the device has two chip enable inputs, it is suited for battery backup applications. Minimum standby power current is drawn by this device when  $\overline{CE1}$  or  $\overline{CE2}$  equals  $V_{CC}$  independently of the other input levels.

Data Retention is guaranteed at a power supply voltage as low as 2V.

The μPD449 is packaged in a standard 24-pin dual-in-line package and is plug-in compatible with 16K EPROMs. The μPD449 is also available in the 24-pin mini-flat (SOP) package.

- FEATURES**
- Single +5V Supply
  - Fully Static Operation – No Clock or Refreshing required
  - TTL Compatible – All Inputs and Outputs
  - Common Data Input and Output Using Three-State Output
  - Two Chip Enable Inputs eliminates need for External Bus Buffers
  - Max Access/Min Cycle Times Down to 150 ns
  - Low Power Dissipation; 18 mA Max Active/10 μA Max Standby/10 μA Max Data Retention
  - Data Retention Voltage – 2V Min
  - Standard 24-Pin Plastic, DIP and Mini-flat (SOP) Packages
  - Plug-in Compatible with 16K EPROMs
  - Operating Temperature Range – 40°C to +85°C
  - For operation at less than 4.5V power supply, contact the NEC Sales Office

**PIN CONFIGURATION**



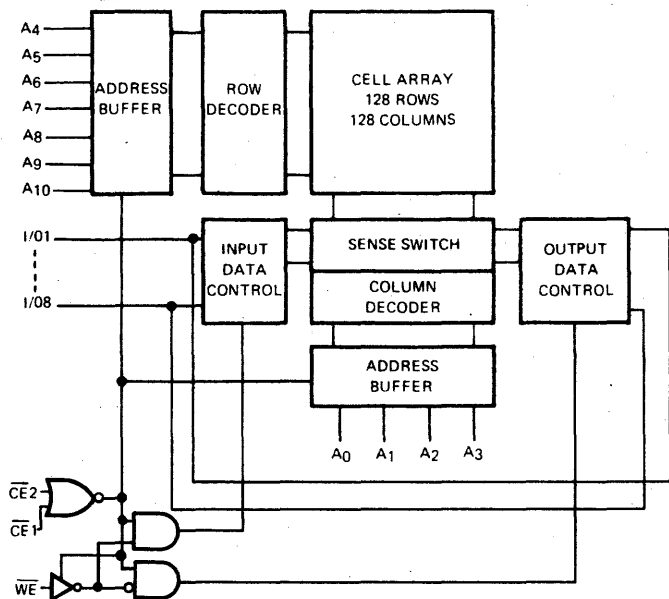
**PIN NAMES**

A0-A10	Address Inputs
$\overline{WE}$	Write Enable
$\overline{CE1}$ - $\overline{CE2}$	Chip Enable Inputs
I/O1-I/O8	Data Input/Output
VCC	Power (+5V)
GND	Ground

**TRUTH TABLE**

$\overline{CE1}$	$\overline{CE2}$	$\overline{WE}$	MODE	I/O	ICC
X	H	X	NOT SELECTED	HZ	STANDBY
H	X	X	NOT SELECTED	HZ	STANDBY
L	L	L	WRITE	DIN	ACTIVE
L	L	H	READ	DOUT	ACTIVE

Rev/1



Supply Voltage ..... 7.0V **ABSOLUTE MAXIMUM RATINGS\***  
 Input or Output Voltage Supplied .....  $-0.3$  to  $V_{CC} + 0.3V$   
 Storage Temperature Range .....  $-55^{\circ}C$  to  $125^{\circ}C$   
 Operating Temperature Range .....  $-40^{\circ}C$  to  $+85^{\circ}C$

$T_a = 25^{\circ}C$

\*COMMENT: Stress above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

$V_{CC} = 5V \pm 10\%$ ,  $T_a = -40^{\circ}C$  to  $+85^{\circ}C$ .

**DC CHARACTERISTICS**

PARAMETER	SYMBOL	LIMITS			UNIT	TEST CONDITIONS
		MIN	TYP	MAX		
Input High Voltage	$V_{IH}$	2.2		$V_{CC} + 0.3$	V	
Input Low Voltage	$V_{IL}$	-0.3		0.8	V	
Input Leakage Current	$I_{LI}$	-1.0		1.0	$\mu A$	$V_{IN} = 0 \sim V_{CC}$
I/O Leakage Current	$I_{LO}$	1.0		1.0	$\mu A$	$V_{CE1}$ or $V_{CE2} = V_{IH}$ or $V_{WE} = V_{IL}$ $V_{I/O} = 0 \sim V_{CC}$
Operating Supply Current	$I_{CCA1}$		①	①	mA	$V_{CE1}$ and $V_{CE2} = V_{IL}$ $I_{I/O} = 0$ MIN TCYCLE
	$I_{CCA2}$		5	10	mA	$V_{CE1}$ and $V_{CE2} = V_{IL}$ $I_{I/O} = 0$ DC CURRENT
	$I_{CCA3}$		30	10	$\mu A$	$V_{CE1}$ and $V_{CE2} = 0.2V$ $V_{IN} = V_{CC} - 0.2V$ or $0.2V$ $I_{I/O} = 0$
Standby Current	$I_{CCS}$			10	$\mu A$	$V_{CE1}$ or $V_{CE2} = V_{CC} - 0.2V$ $V_{IN} = 0 \sim V_{CC}$
Output High Voltage	$V_{OH}$	2.4			V	$I_{OH} = -1.0$ mA
Output Low Voltage	$V_{OL}$			0.4	V	$I_{OL} = 2.0$ mA

NOTE: ①  $\mu PD449$ : 12 mA TYP, 18 mA MAX  
 $\mu PD449-1$ : 18 mA TYP, 26 mA MAX  
 $\mu PD449-2$ : 20 mA TYP, 30 mA MAX  
 $\mu PD449-3$ : 25 mA TYP, 38 mA MAX

MEMORY

CAPACITANCE

$T_a = 25^\circ\text{C}$ ,  $f = 1.0\text{ MHz}$

PARAMETER	SYMBOL	LIMITS		UNIT	TEST CONDITIONS
		MIN	MAX		
Input Capacitance	$C_{IN}$		6	pF	$V_{IN} = 0V$
Input/Output Capacitance	$C_{I/O}$		8	pF	$V_{I/O} = 0V$

AC CHARACTERISTICS

$V_{CC} = 5.0V \pm 10\%$ ,  $T_a = -40^\circ\text{C}$  to  $+85^\circ\text{C}$

READ CYCLE

PARAMETER	SYMBOL	LIMITS								UNIT
		μPD449-3		μPD449-2		μPD449-1		μPD449		
		MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Read Cycle Time	$t_{RC}$	150		200		250		450		ns
Access Time	$t_A$		150		200		250		450	ns
Chip Enable ( $\overline{CE1}$ ) to Output Valid	$t_{CO1}$		150		200		250		450	ns
Chip Enable ( $\overline{CE2}$ ) to Output Valid	$t_{CO2}$		150		200		250		450	ns
Output Hold from Address Change	$t_{OH}$	15		15		15		15		ns
Chip Enable ( $\overline{CE1}$ ) to Output in LZ	$t_{LZ1}$	5		5		5		5		ns
Chip Enable ( $\overline{CE2}$ ) to Output in LZ	$t_{LZ2}$	5		5		5		5		ns
Chip Enable ( $\overline{CE1}$ ) to Output in HZ	$t_{HZ1}$		50		60		80		100	ns
Chip Enable ( $\overline{CE2}$ ) to Output in HZ	$t_{HZ2}$		50		60		80		100	ns

WRITE CYCLE

$V_{CC} = 5.0V \pm 10\%$ ,  $T_a = -40^\circ\text{C}$  to  $+85^\circ\text{C}$

PARAMETER	SYMBOL	LIMITS								UNIT
		μPD449-3		μPD449-2		μPD449-1		μPD449		
		MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Write Cycle Time	$t_{WC}$	150		200		250		450		ns
Chip Enable ( $\overline{CE1}$ ) to End of Write	$t_{CW1}$	120		150		180		210		ns
Chip Enable ( $\overline{CE2}$ ) to End of Write	$t_{CW2}$	120		150		180		210		ns
Address Setup Time	$t_{AW}$	0		0		0		0		ns
Write Pulsewidth	$t_{WP}$	90		120		150		180		ns
Write Recovery Time	$t_{WR}$	0		0		0		0		ns
Write Enable to Output in HZ	$t_{WZ}$		50		60		80		100	ns
Output Active from End of Write	$t_{OW}$	10		10		10		10		ns
Data Valid to End of Write	$t_{DW}$	50		60		80		100		ns
Data Hold Time	$t_{DH}$	0		0		0		0		ns

LOW  $V_{CC}$   
DATA RETENTION

$T_a = -40^\circ\text{C}$  to  $+85^\circ\text{C}$

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP	MAX	
$V_{CC}$ for Data Retention	$V_{CCDR}$	$V_{IN} = 0 \sim V_{CC}$ , $V_{CE1}$ or $V_{CE2} = V_{CC}$	2.0			V
Data Retention Current	$I_{CCDR}$	$V_{CC} = 3.0V$ , $V_{IN} = 0 \sim V_{CC}$ , $V_{CE1}$ or $V_{CE2} = V_{CC}$		0.01	10	μA
Chip Disable to Data Retention Time	$t_{CDR}$		0			ns
Operation Recovery Time	$t_R$		$t_{RC}$			ns

**FEATURES:**

- **Two Fast Access Times:**  
—450nsec  
—350nsec
- **All Inputs TTL Compatible**
- **All Outputs Drive 2 TTL Load Directly**
- **Single +5 Volt Supply with ± 10% Tolerance**
- **Three-State Outputs for Direct Bus Compatibility**
- **Pin Compatible to 2708 EPROM**
- **Fully Static Operation**
- **All Inputs Protected Against Static Charge**
- **100% Burned-In**

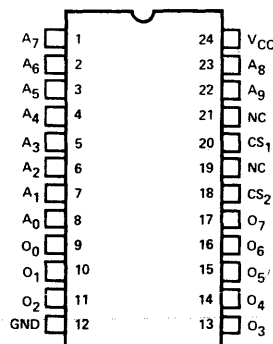
**GENERAL DESCRIPTION:**

The μPD2308A is a 8,192 bit Read Only Memory, utilizing MOS N-Channel silicon gate technology. The device is completely static in operation and is organized as 1024 words by 8 bits and operates from a single +5 volt power supply. All inputs and outputs are fully TTL compatible. It has two programmable chip selects, and Pins 21 and 19 have no connection (NC) to allow full plug-in compatibility with 2708 EPROMs. Programming of the device is done through a custom mask during fabrication.

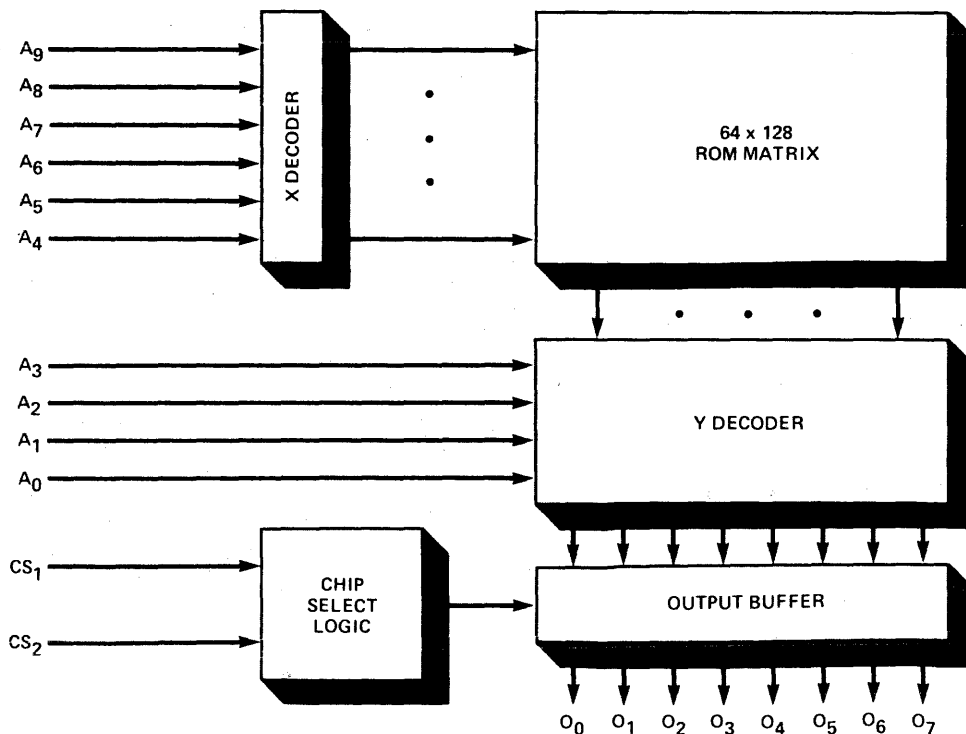
**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
<b>INDUSTRIAL</b>		
μPD 2308A-45PC	450nsec	Molded DIP
μPD 2308A-35PC	350nsec	Molded DIP

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



MEMORY  
NEC Electronics

**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

**D.C. CHARACTERISTICS** (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IL</sub>	Input "Low" Voltage		-0.5		0.8	V
V <sub>IH</sub>	Input "High" Voltage		2.0		V <sub>CC</sub> + 1	V
I <sub>IL</sub>	Input Load Current	V <sub>IN</sub> = 0 to +5.50V			10	μA
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = +3.2mA			0.40	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -200μA	2.4			V
I <sub>LO</sub>	Output Leakage Current	Chip Disabled V <sub>OUT</sub> = +0.4V to V <sub>CC</sub>			10	μA
I <sub>CC</sub>	Power Supply Current	All inputs +5.50V Outputs Unloaded		60	90	mA

**A.C. CHARACTERISTICS** (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	μPD2308A-35		μPD2308A-45		Units
		Min.	Max.	Min.	Max.	
t <sub>ACC</sub>	Access Time		350		450	nsec
t <sub>CO</sub>	Output Enable Time		150		150	nsec
t <sub>DF</sub>	Output Disable Time		100		100	nsec
t <sub>OH</sub>	Output Hold Delay	20		20		nsec
C <sub>I</sub>	Input Capacitance		7		7	pf
C <sub>O</sub>	Output Capacitance		10		10	pf

**A.C. TEST CONDITIONS**

Input rise and fall times (t<sub>r</sub>, t<sub>f</sub>) = 20 nsec  
 Timing Measurement Reference Levels:  
 Input and output high levels (V<sub>IH</sub>, V<sub>OH</sub>) = 2.0 Volts  
 Input and output low levels (V<sub>IL</sub>, V<sub>OL</sub>) = 0.8 Volts  
 Output load = 1 series 74 TTL + 100 pf

**CUSTOM PROGRAMMING INSTRUCTIONS**

**BIT PATTERN SUBMITTAL OPTIONS:**

The customer's unique bit pattern can be submitted in a convenient method that is easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:  
 1. One programmed 2708 EPROM

**BIT PATTERN VERIFICATION:**

For customer verification of the submitted bit patterns, several alternatives are also available. The following is found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. One programmed 2708 EPROM	Customer sends NEC one additional erased 2708. NEC programs the spare 2708 with the data from the programmed 2708, and returns to customer for verification.

**FEATURES:**

- **Two Fast Access Times:**  
—450nsec  
—350nsec
- **All Inputs TTL Compatible**
- **Single +5 V ± 10% Supply**
- **Three-State Outputs for Direct Bus Compatibility**
- **Pin Compatible to 2708 and 2716 EPROMs**
- **Fully Static Operation**
- **Inputs Static Charge Protected**
- **100% Burned-In**

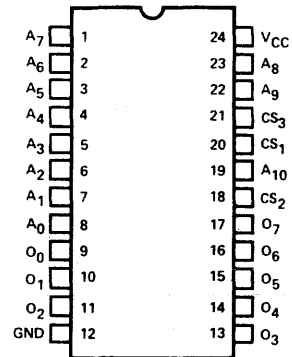
**GENERAL DESCRIPTION:**

The μPD2316E is a 16,384 bit Read Only Memory utilizing MOS N-Channel silicon gate technology. The device is completely static in operation and is organized as 2048 words by 8 bits and operates from a single +5 volt power supply. All inputs and outputs are fully TTL compatible. It has three programmable chip select inputs and three state outputs that allow memory expansion to 16,384 words by 8 bits without the use of external logic. Programming of the device is accomplished by a custom mask during fabrication. Pinout is compatible with 2708 and 2716 EPROMs.

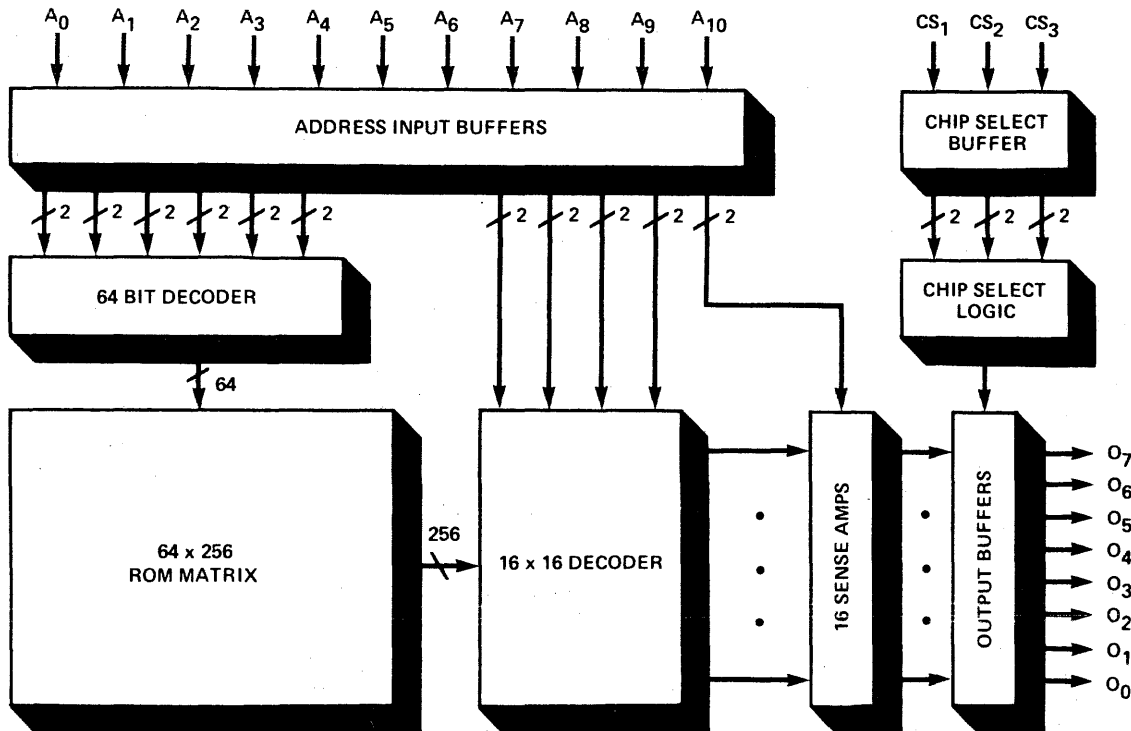
**ORDERING INFORMATION**

PART NUMBER INDUSTRIAL	ACCESS TIME	PACKAGE
μPD 2316E-45PC	450nsec	Molded DIP
μPD 2316E-35PC	350nsec	Molded DIP

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



MEMORY  
NEC Electronics

**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

**D.C. CHARACTERISTICS** (T<sub>a</sub> = 0°C to +70°C, V<sub>CC</sub> = +5±10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IL</sub>	Input "Low" Voltage		-0.5		0.8	V
V <sub>IH</sub>	Input "High" Voltage		2.0		V <sub>CC</sub> + 1	V
I <sub>IL</sub>	Input Load Current	V <sub>IN</sub> = 0 to +5.50V			10	μA
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = +3.2mA			0.40	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -200μA	2.4			V
I <sub>LO</sub>	Output Leakage Current	Chip Disabled, V <sub>OUT</sub> = +0.4V to V <sub>CC</sub>			10	μA
I <sub>CC</sub>	Power Supply Current	All inputs +5.50V, Outputs Unloaded		60	90	mA

**A.C. CHARACTERISTICS** (T<sub>a</sub> = 0°C to +70°C, V<sub>CC</sub> = +5±10%)

Symbol	Parameter	μPD2316E-35		μPD2316E-45		Units
		Min.	Max.	Min.	Max.	
t <sub>ACC</sub>	Access Time		350		450	nsec
t <sub>CO</sub>	Output Enable Time		150		150	nsec
t <sub>DF</sub>	Output Disable Time		100		100	nsec
t <sub>OH</sub>	Output Hold Delay	20		20		nsec
C <sub>I</sub>	Input Capacitance		7		7	pf
C <sub>O</sub>	Output Capacitance		10		10	pf

**A.C. TEST CONDITIONS**

Input rise and fall times (t<sub>r</sub>, t<sub>f</sub>) = 20 nsec  
 Timing Measurement Reference Levels:  
 Input and output high levels (V<sub>IH</sub>, V<sub>OH</sub>) = 2.0 Volts  
 Input and output low levels (V<sub>IL</sub>, V<sub>OL</sub>) = 0.8 Volts  
 Output load = 1 series 74 TTL + 100 pf

**CUSTOM PROGRAMMING INSTRUCTIONS**

**BIT PATTERN SUBMITTAL OPTIONS:**

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:  
 1. One programmed 2716 EPROM  
 2. Two programmed 2708 EPROMs

**BIT PATTERN VERIFICATION:**

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. One programmed 2716 EPROM	Customer sends NEC one additional erased 2716. NEC programs the spare 2716 with the data extracted from the programmed 2716, and returns to the customer for verification.
2. Two programmed 2708's EPROMs	Customer sends NEC two additional erased 2708's. NEC programs the spare 2708's with the pattern data extracted from the programmed 2708's and returns to the customer for verification.

**FEATURES:**

- Three Fast Access Times  
-450nsec  
-350nsec  
-300nsec
  - All Inputs TTL Compatible
  - Outputs Drive 2 TTL Loads Directly
  - Single +5V ± 10% Supply
  - Two Programmable Chip Selects
  - Both JEDEC Pinouts Available
- |          |        |        |
|----------|--------|--------|
|          | PIN 18 | PIN 21 |
| μPD2332A | A11    | CS2    |
| μPD2332B | CS2    | A11    |
- Pin Compatible to 2716 and 2732 EPROMS
  - Fully Static Operation
  - All Inputs Protected Against Static Charge
  - 100% Burned-In

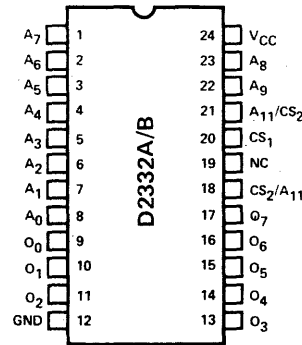
**ORDERING INFORMATION**

PART NUMBER	PIN 18	PIN 21	ACCESS TIME	PACKAGE
μPD 2332A-45PC	A11	CS2	450ns	Molded DIP
μPD 2332A-35PC	A11	CS2	350ns	Molded DIP
μPD 2332A-30PC	A11	CS2	300ns	Molded DIP
μPD 2332B-45PC	CS2	A11	450ns	Molded DIP
μPD 2332B-35PC	CS2	A11	350ns	Molded DIP
μPD 2332B-30PC	CS2	A11	300ns	Molded DIP

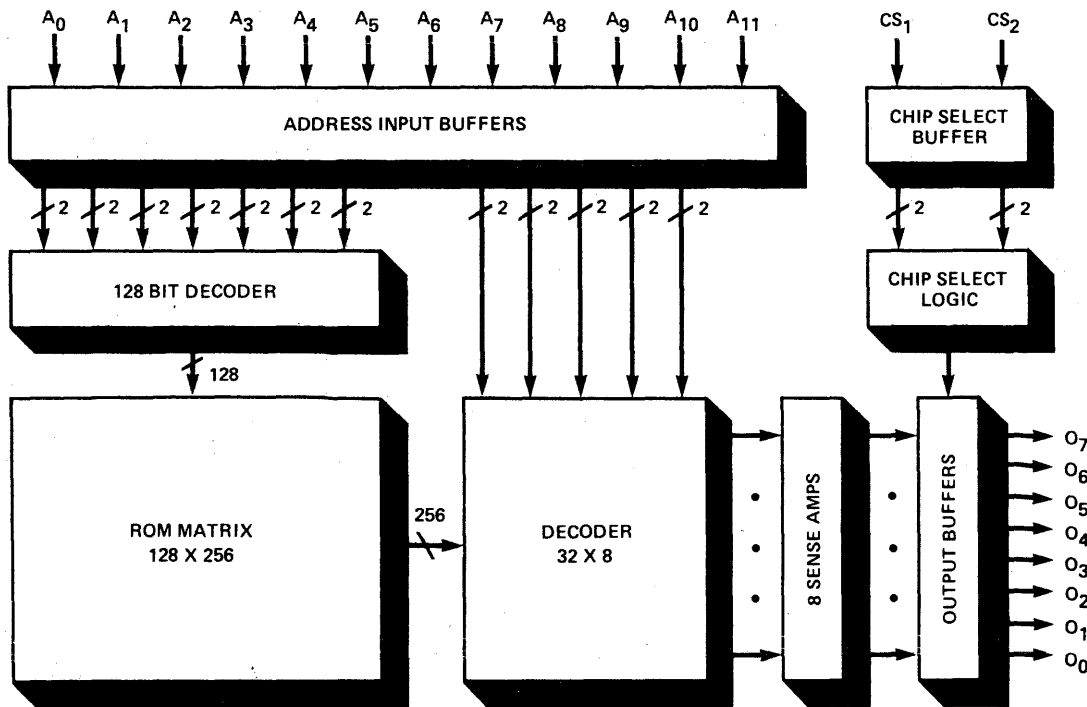
**GENERAL DESCRIPTION:**

The μPD2332A/B is a 32,768 bit fully static Read Only Memory utilizing MOSN-Channel silicon gate ion implanted technology. It is organized 4096 words by 8 bits and operates from a single +5V ± 10% power supply. All inputs are TTL compatible. It has two programmable chip selects and three-state outputs for direct bus compatibility. It is unique in that both JEDEC standard pin configurations are available.

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**





## ABSOLUTE MAXIMUM RATINGS

**μPD2332**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

### D.C. CHARACTERISTICS (T<sub>a</sub> = 0° to +70°C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IL</sub>	Input "Low" Voltage		-0.5		0.8	V
V <sub>IH</sub>	Input "High" Voltage		2.0		V <sub>CC</sub> + 1	V
I <sub>IL</sub>	Input Load Current	V <sub>IN</sub> = 0 to 5.5V			10	μA
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = +3.2mA			0.40	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -200μA	2.4			V
I <sub>LO</sub>	Output Leakage Current	Chip Disabled V <sub>OUT</sub> = +0.4V to V <sub>CC</sub>			10	μA
I <sub>CC</sub>	Power Supply Current	All inputs +5.5V Output Unloaded		60	90	mA

### A.C. CHARACTERISTICS (T<sub>a</sub> = 0° to +70°C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	μPD2332A/B-30		μPD2332A/B-35		μPD2332A/B-45		Unit
		Min.	Max.	Min.	Max.	Min.	Max.	
t <sub>ACC</sub>	Access Time		300		350		450	nsec
t <sub>CO</sub>	Output Enable Time		150		150		150	nsec
t <sub>DF</sub>	Output Disable Time		100		100		100	nsec
t <sub>OH</sub>	Output Hold Delay	20		20		20		nsec
C <sub>I</sub>	Input Capacitance		7		7		7	pf
C <sub>O</sub>	Output Capacitance		10		10		10	pf

### A.C. TEST CONDITIONS

Input rise and fall times (t<sub>r</sub>, t<sub>f</sub>) = 20 nsec

Timing Measurement Reference Levels:

Input and output high levels (V<sub>IH</sub>, V<sub>OH</sub>) = 2.0 Volts

Input and output low levels (V<sub>IL</sub>, V<sub>OL</sub>) = 0.8 Volts

Output load = 1 series 74 TTL + 100 pf

### CUSTOM PROGRAMMING INSTRUCTIONS

#### BIT PATTERN SUBMITTAL OPTIONS:

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:

1. One programmed 2732 EPROM
2. Two programmed 2716 EPROMs

#### BIT PATTERN VERIFICATIONS:

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. One programmed 2732 EPROM	Customer sends NEC one additional erased 2732. NEC programs the spare 2732 with the data from the programmed 2732, and returns to customer for verification.
2. Two programmed 2716 EPROMs	Customer sends NEC two additional erased 2716's. NEC programs the spare 2716's with the data from the programmed EPROMs and returns to customer for verification.

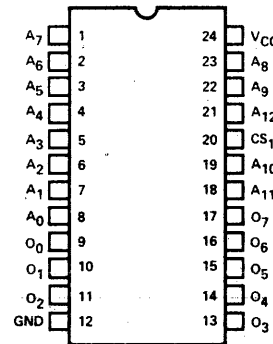
**FEATURES:**

- **Three Fast Access Times**  
-450nsec  
-350nsec  
-300nsec
- **All Inputs and Outputs TTL Compatible**
- **Single +5V ± 10% Power Supply**
- **One Programmable Chip Select**
- **Three-State Outputs for Direct Bus Compatibility**
- **Pin Compatible to 2716, 2732, and 2564 EPROMS**
- **Fully Static Operation**
- **All Inputs Protected Against Static Charge**
- **100% Burned-In**

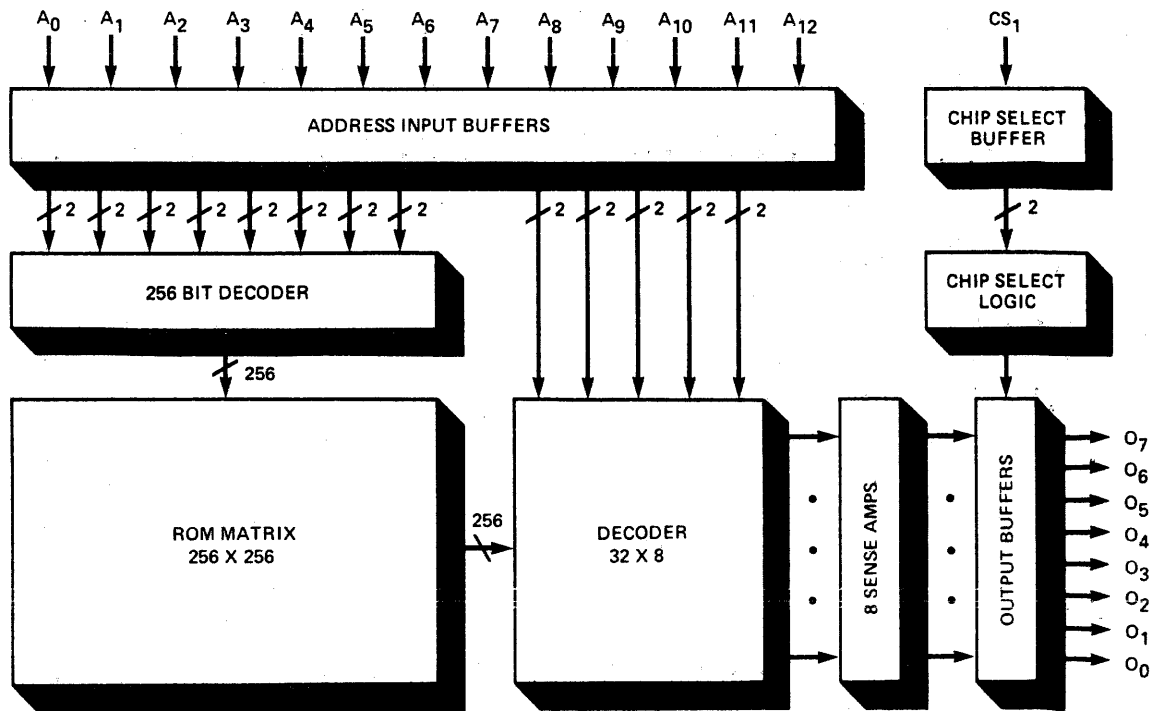
**GENERAL DESCRIPTION:**

The μPD2364 is a 65,536 bit Read Only Memory utilizing MOS N-Channel silicon gate technology. The device is completely static in operation, organized as 8192 words by 8 bits and operates from a single +5V power supply. All inputs and outputs are fully TTL compatible. It has one programmable chip select input and three-state outputs that allow memory expansion to 16,384 words by 8 bits without the use of external logic. Programming of the device is accomplished by a custom mask during fabrication. Pinout is compatible with 2716, 2732, and 2564 EPROMs.

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



NEC Electronics  
MEMORY

**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

**D.C. CHARACTERISTICS** (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IL</sub>	Input "Low" Voltage		-0.5		0.8	V
V <sub>IH</sub>	Input "High" Voltage		2.0		V <sub>CC</sub> + 1	V
I <sub>IL</sub>	Input Load Current	V <sub>IN</sub> = 0 to 5.5V			10	μA
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = +2.1mA			0.4	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -400μA	2.4			V
I <sub>LO</sub>	Output Leakage Current	Chip Disabled V <sub>OUT</sub> = +0.4V to V <sub>CC</sub>			10	μA
I <sub>CC</sub>	Power Supply Current	All inputs +5.5V Output Disabled		80	140	mA

**A.C. CHARACTERISTICS** (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	μPD2364-30		μPD2364-35		μPD2364-45		Unit
		Min.	Max.	Min.	Max.	Min.	Max.	
t <sub>ACC</sub>	Access Time		300		350		450	nsec
t <sub>CO</sub>	Output Enable Time		120		150		150	nsec
t <sub>DF</sub>	Output Disable Time		120		150		150	nsec
t <sub>OH</sub>	Output Hold Delay	20		20		20		nsec
C <sub>I</sub>	Input Capacitance		10		10		10	pf
C <sub>O</sub>	Output Capacitance		15		15		15	pf

**A.C. TEST CONDITIONS**

Input rise and fall times (t<sub>r</sub>, t<sub>f</sub>) = 20 nsec  
 Timing Measurement Reference Levels:  
 Input and output high levels (V<sub>IH</sub>, V<sub>OH</sub>) = 2.0 Volts  
 Input and output low levels (V<sub>IL</sub>, V<sub>OL</sub>) = 0.8 Volts  
 Output load = 1 series 74 TTL + 100 pf

**CUSTOM PROGRAMMING INSTRUCTIONS**

**BIT PATTERN SUBMITTAL OPTIONS:**

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:  
 1. One programmed 2764 EPROM  
 2. Two programmed 2732 EPROMs

**BIT PATTERN VERIFICATIONS:**

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. One programmed 2764 EPROM	Customer sends NEC one additional erased 2764. NEC programs the spare 2764 with the data from the programmed 2764, and returns to customer for verification.
2. Two programmed 2732 EPROMs	Customer sends NEC two additional erased 2732's. NEC programs the spare 2732's with the data from the programmed EPROMs and returns to customer for verification.

**FEATURES:**

- **Two Fast Access Times:**  
-250nsec—μPD2364E-25  
-200nsec—μPD2364E-20
- **All Inputs and Outputs are fully TTL Compatible**
- **Three-State Output for Direct Bus Compatibility**
- **Single +5V ± 5% Power Supply**
- **Three Programmable Chip Selects**
- **All Inputs Protected Against Static Charge**
- **100% Burned-In**

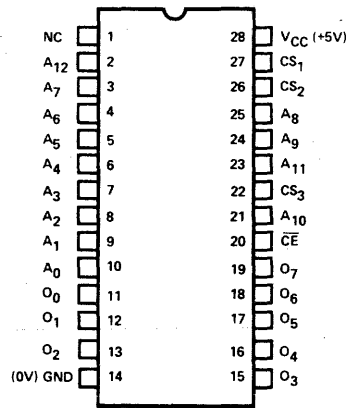
**GENERAL DESCRIPTION:**

The μPD2364E-20/2364E-25 is a 65,536 bit Read Only utilizing NMOS silicon gate technology. The device is static in operation, organized as 8192 words by 8 bits and operates from a single +5 Volt power supply. The device has three-state outputs and all inputs and outputs are fully TTL compatible. The chip select pins are mask programmable and can be specified by selecting "1", "0" and "Don't Care" Data. Pinout is compatible with 2764 EPROMs.

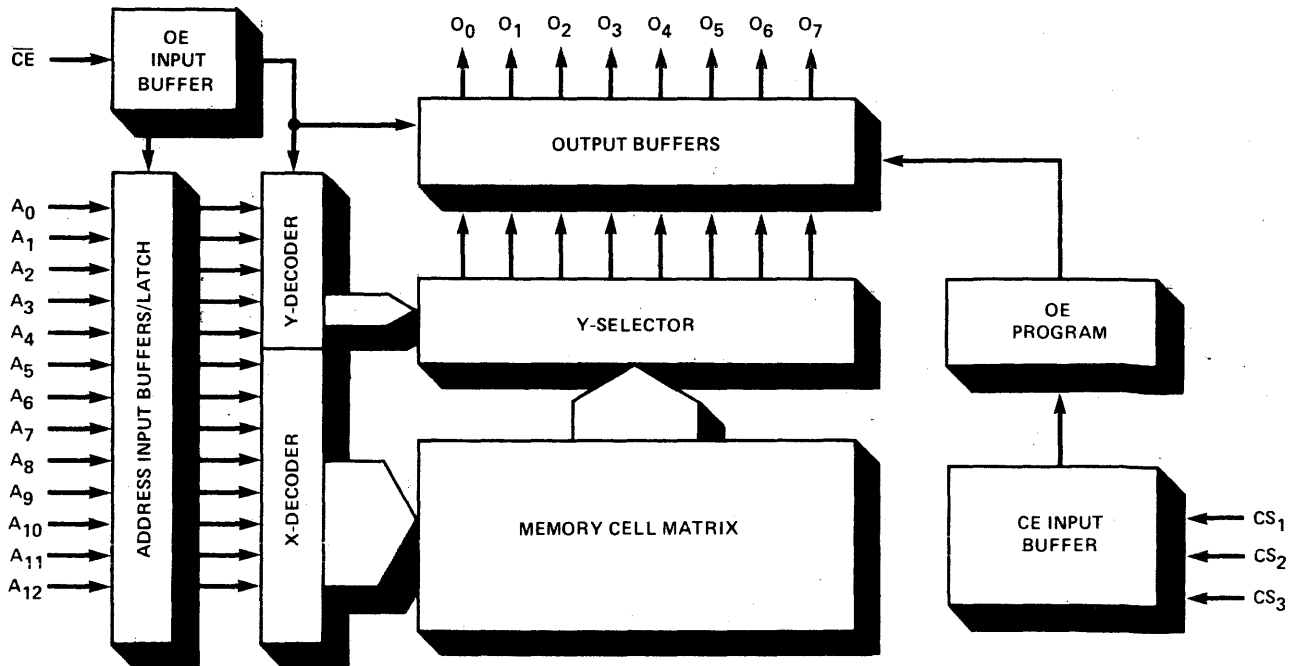
**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
μPD 2364E-25PC	250nsec	Molded 28-PIN DIP
μPD 2364E-20PC	200nsec	Molded 28-PIN DIP

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



MEMORY  
NEC Electronics

## ABSOLUTE MAXIMUM RATINGS

**μPD2364E**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

### D.C. CHARACTERISTICS (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5 ± 5%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IH</sub>	Input "High" Voltage		+2.1		V <sub>CC</sub> + 1.0	V
V <sub>IL</sub>	Input "Low" Voltage		-0.5		+0.7	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -400 μA	+2.4			V
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = +2.1 mA			+0.4	V
I <sub>LIH</sub>	Input Leakage Current High	V <sub>I</sub> = V <sub>CC</sub>			+10	μA
I <sub>LIL</sub>	Input Leakage Current Low	V <sub>I</sub> = 0V			-10	μA
I <sub>LOH</sub>	Output Leakage Current High	V <sub>O</sub> = V <sub>CC</sub> , Chip Deselected			+10	μA
I <sub>LOL</sub>	Output Leakage Current Low	V <sub>O</sub> = 0V, Chip Deselected			-10	μA
I <sub>CC1</sub>	Supply Voltage	CE = V <sub>IL</sub>		45	85	mA
I <sub>CC2</sub>	Supply Voltage	CE = V <sub>IH</sub> , Standby Mode		12	20	mA

### A.C. CHARACTERISTICS (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5 ± 5%)

Symbol	Parameter	Test Condition	D2364E-25			D2364E-20			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
t <sub>ACC</sub>	Access Time	Input Voltage tr, tf = 20ns Timing Reference Levels: Input = 0.8V & 2.0V Output = 0.8V & 2.0V Load 1 TTL + 100 pF			250			200	ns
t <sub>CE</sub>	Chip Enable Access Time				250			200	ns
t <sub>OE</sub>	OE <sub>1</sub> to OE <sub>3</sub> Output On Time		10		110	10		100	ns
t <sub>OH</sub>	Output Hold Time		0			0			ns
t <sub>DF</sub>	Output Disable Time		0		100	0		90	ns
C <sub>I</sub>	Input Capacitance	f = 1MHz			10			10	pF
C <sub>O</sub>	Output Capacitance	f = 1MHz			15			15	pF

## CUSTOM PROGRAMMING INSTRUCTIONS

### BIT PATTERN SUBMITTAL OPTIONS:

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:

1. One programmed 2764 EPROM
2. Two programmed 2732 EPROMs

### BIT PATTERN VERIFICATIONS:

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

#### CUSTOM PATTERN SUBMITTED VIA:

1. One programmed 2764 EPROM
2. Two programmed 2732 EPROMs

#### VERIFICATION ROUTINE

Customer sends NEC one additional erased 2764. NEC programs the spare 2764 with the data from the programmed 2764, and returns to customer for verification.

Customer sends NEC two additional erased 2732's. NEC programs the spare 2732's with the data from the programmed EPROMs and returns to customer for verification.

**FEATURES:**

- Fast Access Time: 250nsec Max.
- All Inputs and Outputs are TTL Compatible
- Single +5 Volt ± 10% Supply
- Three-State Outputs for Direct Bus Compatibility
- Edge Enabled Operation
- Programmable Chip Select for Memory Expansion
- Low-Power Standby Mode
- All Inputs Protected Against Static Charge
- 100% Burned-In

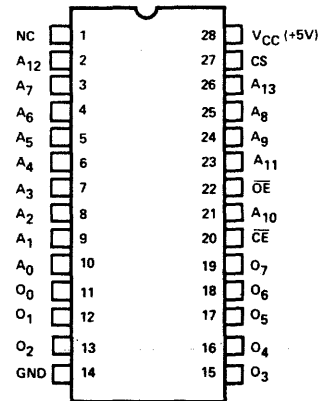
**GENERAL DESCRIPTION:**

The μPD23128 is a 131,072 bit edge enabled Read Only Memory, utilizing MOS N-Channel silicon gate, ion implanted technology. The device is organized 16,384 words by 8 bits and operates from a single +5 volt supply. All inputs and outputs are fully TTL compatible and the device has one programmable chip select, with three-state outputs that allow memory expansion without the use of external logic. Programming of the device is accomplished during the fabrication process. Pinout is compatible with 27128 EPROMs.

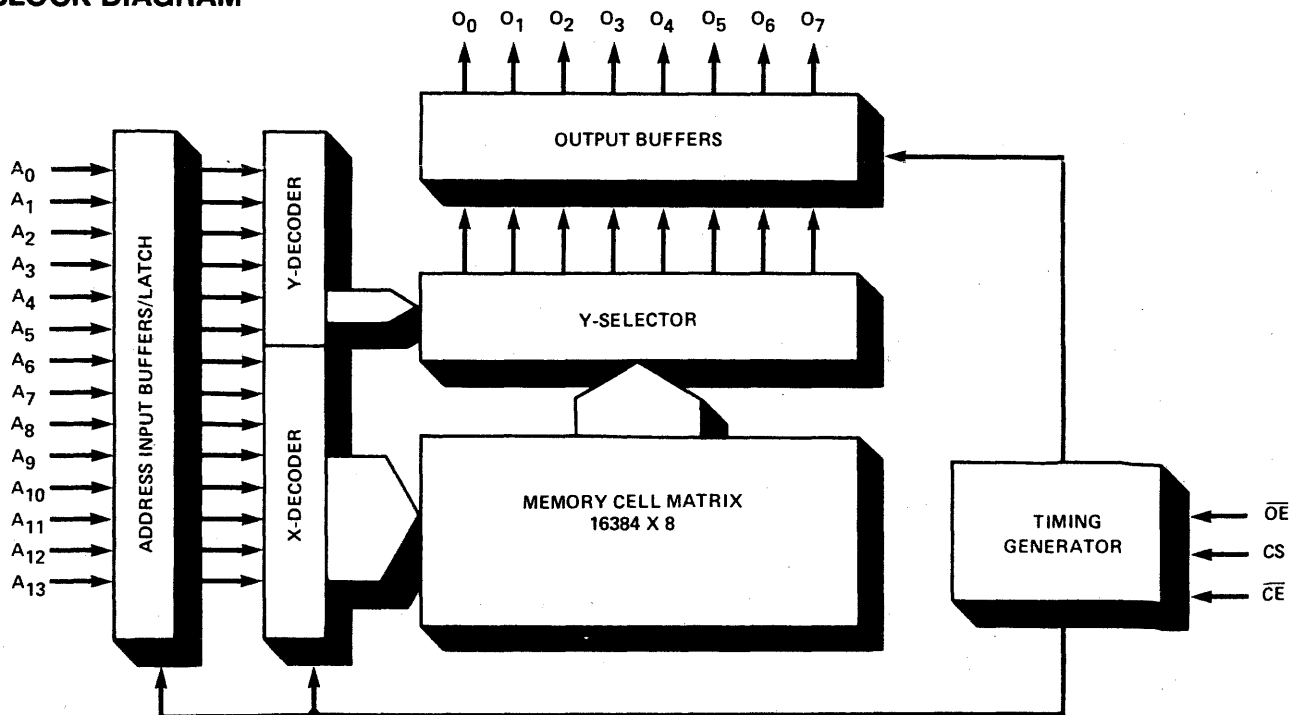
**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
μPD 23128-25PC	250nsec	Plastic DIP

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



MEMORY  
NEC Electronics

**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

**D.C. CHARACTERISTICS** (T<sub>a</sub> = 0°C to +70°C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IH</sub>	Input "High" Voltage		+2.0		V <sub>CC</sub> + 1.0	V
V <sub>IL</sub>	Input "Low" Voltage		-0.5		+0.8	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -400μA	+2.4			V
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = +3.2mA			+0.4	V
I <sub>LIH</sub>	Input Leakage Current	V <sub>I</sub> = V <sub>CC</sub>			+10	μA
I <sub>LIL</sub>	Input Leakage Current	V <sub>I</sub> = 0V			-10	μA
I <sub>LOH</sub>	Output Leakage Current	V <sub>O</sub> = V <sub>CC</sub> , Chip Deselected			+10	μA
I <sub>LOL</sub>	Output Leakage Current	V <sub>O</sub> = 0V, Chip Deselected			-10	μA
I <sub>CC1</sub>	Power Supply Current	t <sub>CYC</sub> = 350ns		+25	+40	mA
I <sub>CC2</sub>	Power Supply Current	Standby Mode		+7	+15	mA

**A.C. CHARACTERISTICS** (T<sub>a</sub> = 0°C to +70°C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Units	
t <sub>CYC</sub>	Cycle Time	Output Load = 1TTL + 100pF Input Voltage t <sub>r</sub> , t <sub>f</sub> = 20 ns Timing Reference Levels: Input Voltage = 0.8V & 2.0V Output Voltage = 0.8V & 2.0V	350			ns	
t <sub>AS</sub>	Address Setup Time		0			ns	
t <sub>AH</sub>	Address Hold Time		60			ns	
t <sub>CE</sub>	Chip Enable Access Time				250	ns	
t <sub>OE</sub>	Output Enable Access Time				120	ns	
t <sub>ACC</sub>	Access Time <sup>(1)</sup>				250	ns	
t <sub>DF</sub>	Output Disable Time			0		70	ns
t <sub>CC</sub>	Chip Enable Off Time			50			ns
C <sub>I</sub>	Input Capacitance	f = 1MHz			10	pF	
C <sub>O</sub>	Output Capacitance	f = 1MHz			15	pF	

**CUSTOM PROGRAMMING INSTRUCTIONS**

**BIT PATTERN SUBMITTAL OPTIONS:**

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:

- One programmed 27128 EPROM
- Two programmed 2764 EPROMs
- Four programmed 2732 EPROMs

**BIT PATTERN VERIFICATIONS:**

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. One programmed 27128 EPROM	Customer sends NEC one additional erased 27128. NEC programs the spare 27128 with the data from the programmed 27128, and returns to customer for verification.
2. Two programmed 2764 EPROMs	Customer sends NEC two additional erased 2764's. NEC programs the spare 2764's with the data from the programmed 2764's and returns to the customer for verification.
3. Four programmed 2732 EPROMs	Customer sends NEC four additional erased 2732's. NEC programs the spare 2732's with the data from the programmed 2732's and returns to the customer for verification.

**FEATURES:**

- Fast Access: 250nsec Max.
- All Inputs and Outputs are TTL Compatible
- Single +5 Volt  $\pm$  10% Supply
- Three-State Outputs for Direct Bus Compatibility
- Edge Enabled Operation
- Low-Power Standby Mode
- One Programmable Chip Select for Memory Expansion
- All Inputs Protected Against Static Charge
- 100% Burned-In

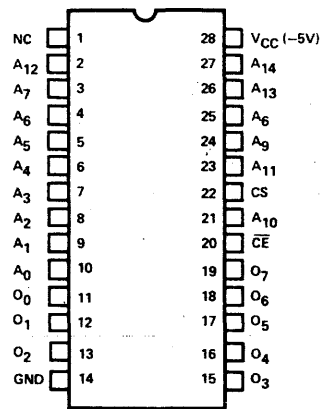
**GENERAL DESCRIPTION:**

The  $\mu$ PD23256 is a 262,144 bit edge enabled Read Only Memory, utilizing MOS N-Channel silicon gate, ion implanted technology. The device is organized 32,768 words by 8-bits and operates from a single +5 Volt supply. All inputs and outputs are fully TTL compatible with one programmable chip select, and the device has three-state outputs that allow memory expansion without the use of external logic. Programming of the device is accomplished during the fabrication process and the pinout is compatible with 27128 EPROMs.

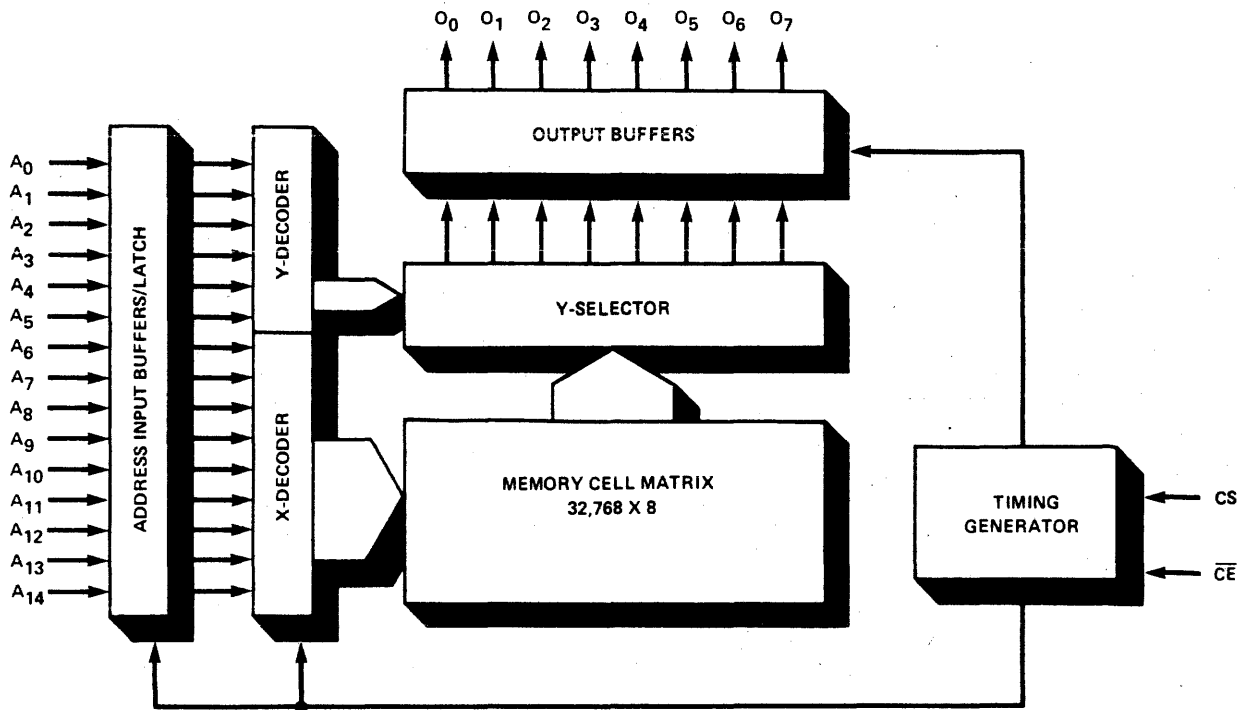
**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
$\mu$ PD 23256-25PC	250nsec	Plastic DIP

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



MEMORY NEC Electronics



## ABSOLUTE MAXIMUM RATINGS

**μPD23256**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

## D.C. CHARACTERISTICS (T<sub>a</sub> = 0°C to +70°C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IH</sub>	Input "High" Voltage		+2.0		V <sub>CC</sub> + 1.0	V
V <sub>IL</sub>	Input "Low" Voltage		-0.5		+0.8	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -400μA	+2.4			V
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OH</sub> = +3.2mA			+0.4	V
I <sub>LIH</sub>	Input Leakage Current High	V <sub>I</sub> = V <sub>CC</sub>			+10	μA
I <sub>LIL</sub>	Input Leakage Current Low	V <sub>I</sub> = 0V			-10	μA
I <sub>LOH</sub>	Output Leakage Current High	V <sub>O</sub> = V <sub>CC</sub> , Chip Deselected			+10	μA
I <sub>LOL</sub>	Output Leakage Current Low	V <sub>O</sub> = 0V, Chip Deselected			-10	μA
I <sub>CC1</sub>	Power Supply Current	t <sub>CYC</sub> = 350ns		+25	+40	mA
I <sub>CC2</sub>	Power Supply Current	Standby Mode		+8	+15	mA

## A.C. CHARACTERISTICS (T<sub>a</sub> = 0°C to +70°C, V<sub>CC</sub> = +5 ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Units	
t <sub>CYC</sub>	Cycle Time	Input Rise/Fall Times: 20ns Timing Reference Levels: Input Voltage = 0.8V & 2.0V Output Voltage = 0.8V & 2.0V Output Load = 1 TTL + 100pF	350			ns	
t <sub>AS</sub>	Address Setup Time		0			ns	
t <sub>AH</sub>	Address Hold Time		60			ns	
t <sub>ACC</sub>	Access Time <sup>(1)</sup>				250	ns	
t <sub>CE</sub>	Chip Enable Access Time				250	ns	
t <sub>OE</sub>	Output Enable Access Time				120	ns	
t <sub>DF</sub>	Output Disable Time			0		70	ns
t <sub>CC</sub>	Chip Enable Off Time			50			ns
C <sub>I</sub>	Input Capacitance		f = 1MHz			10	pF
C <sub>O</sub>	Output Capacitance	f = 1MHz			15	pF	

## CUSTOM PROGRAMMING INSTRUCTIONS

### BIT PATTERN SUBMITTAL OPTIONS:

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:

- Two programmed 27128 EPROMs
- Four programmed 2764 EPROMs

### BIT PATTERN VERIFICATIONS:

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. Two programmed 27128 EPROMs	Customer sends NEC two additional erased 27128's. NEC programs the spare 27128's with the data from the programmed 27128's, and returns to the customer for verification.
2. Four programmed 2764 EPROMs	Customer sends NEC four additional erased 2764's. NEC programs the spare 2764's with the data from the programmed 2764's, and returns to the customer for verification.

**FEATURES:**

- **Two Fast Access Times:**  
— 350nsec- $\mu$ PD231000-35  
— 300nsec- $\mu$ PD231000-30
- **All Inputs and Outputs are Fully TTL Compatible**
- **Single +5 Volt Power Supply**
- **Three-State Outputs for Direct Bus Compatibility**
- **Edge Enabled Operation**
- **Low Power Standby Mode**
- **All Inputs Protected Against Static Charge**
- **100% Burned-In**

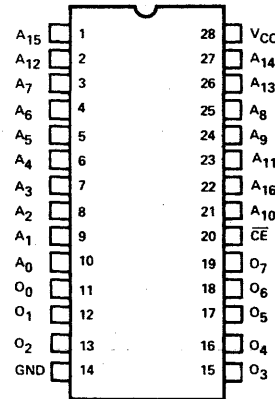
**GENERAL DESCRIPTION:**

The  $\mu$ PD231000 is a 1,048,576 bit Read Only Memory, utilizing NMOS silicon gate technology. The device is edge enabled, organized as 131,072 words by 8 bits and operates from a single +5 Volt power supply. All inputs and outputs are fully TTL compatible. The device has three state outputs for direct bus interface. The device has a low-power stand-by mode and is available in two speed versions (300 and 350nsec). Programming of the device is accomplished by a custom mask during the fabrication process.

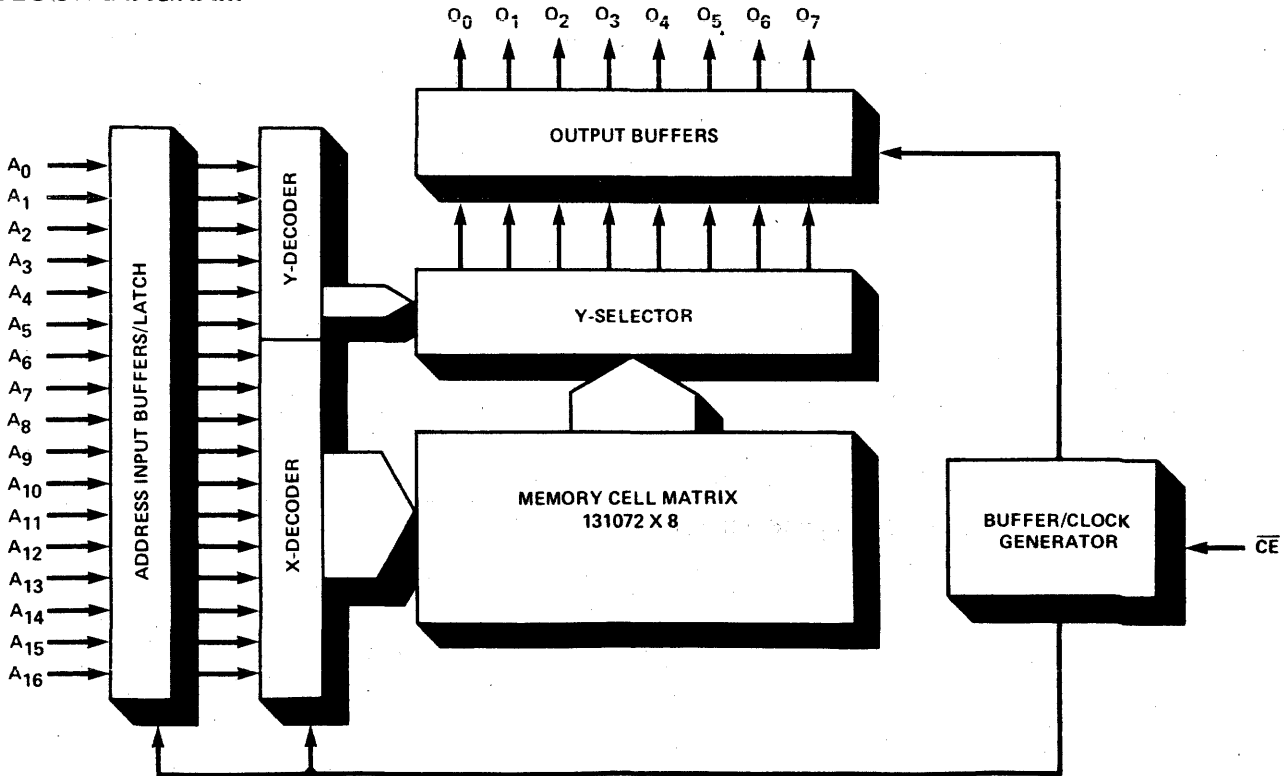
**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
$\mu$ PD 231000-35PC	350nsec	Plastic DIP
$\mu$ PD 231000-30PC	300nsec	Plastic DIP

**PIN CONFIGURATION (TOP VIEW)**



**BLOCK DIAGRAM**



MEMORY  
NEC Electronics

**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.5 to +7	V
V <sub>I</sub>	Input Voltage	-0.5 to +7	V
V <sub>O</sub>	Output Voltage	-0.5 to +7	V
T <sub>opt</sub>	Operating Temperature	-0 to +70	°C
T <sub>stg</sub>	Storage Temperature	-65 to +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

**D.C. CHARACTERISTICS** (T<sub>a</sub> = 0° to +70° C, V<sub>CC</sub> = +5V ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V <sub>IH</sub>	Input "High" Voltage		+2.0		V <sub>CC</sub> +1.0	V
V <sub>IL</sub>	Input "Low" Voltage		-0.5		+0.8	V
V <sub>OH</sub>	Output "High" Voltage	I = -400μA	+2.4			V
V <sub>OL</sub>	Output "Low" Voltage	I = +3.2mA			+0.4	V
I <sub>LIH</sub>	Input Leakage Current High	V = V <sub>CC</sub>			+10	μA
I <sub>LIL</sub>	Input Leakage Current Low	V = 0V			-10	μA
I <sub>LOH</sub>	Output Leakage Current High	V = V <sub>CC</sub> , Chip Deselected			+10	μA
I <sub>LOL</sub>	Output Leakage Current Low	V = 0V, Chip Deselected			-10	μA
I <sub>CC1</sub>	Power Supply Current	Operating Mode (Average) t <sub>CYC</sub> = 450ns		35	50	mA
I <sub>CC2</sub>	Power Supply Current	Standby Mode		8	15	mA

**A.C. CHARACTERISTICS** (T<sub>a</sub> = 0° C to +70° C, V<sub>CC</sub> = +5V ± 10%)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Units
t <sub>CYC</sub>	Cycle Time	Input Voltage t <sub>r</sub> , t <sub>f</sub> = 20 ns Timing Reference Levels: Input Voltage = 0.8V & 2.0V Output Voltage = 0.8V & 2.0V Output Load = 1TTL + 100pF	450			ns
t <sub>AS</sub>	Address Setup Time		0			ns
t <sub>AH</sub>	Address Hold Time		100			ns
t <sub>ACC</sub>	Access Time				350	ns
t <sub>CE</sub>	Chip Enable Access Time				350	ns
t <sub>CC</sub>	Chip Enable Off Time		70			ns
C <sub>I</sub>	Input Capacitance	f = 1MHz			10	pF
C <sub>O</sub>	Output Capacitance	f = 1MHz			15	pF

**CUSTOM PROGRAMMING INSTRUCTIONS**

**BIT PATTERN SUBMITTAL**

**OPTIONS:**

The customer's unique bit pattern can be submitted in a convenient method that is easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:

1. Eight programmed 27128 EPROMs

**BIT PATTERN VERIFICATIONS:**

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. Eight programmed 27128 EPROMs	Customer sends NEC eight additional erased 27128's. NEC programs the spare 27128's with the data from the programmed 27128's, and returns to the customer for verification.

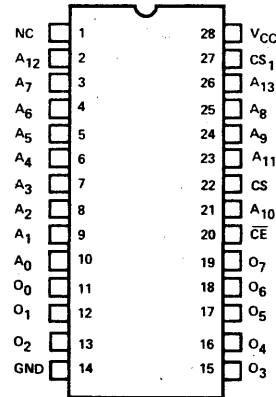
**FEATURES:**

- 16,384 Words by 8 Bits
- Access Time: 5μsec  $V_{CC} = 5V \pm 10\%$
- Low Power Consumption—CMOS Technology
- One Mask Programmable Chip Select
- Three State Output for Direct Bus Interface
- Built for Direct Interface with 8048/8048 μproc.
- Single Power Supply (3.5-5.5V)
- 100% Burned-In

**GENERAL DESCRIPTION:**

The μPD23C128 is a 131,072 bit Read Only Memory utilizing low power CMOS technology. The device is mask programmable and is organized 16,384 words, by 8 bits/word and operates from a single 3.5-5.5 volt supply. The device has one mask programmable chip select input for simple memory expansion and has three-state output for direct bus interface.

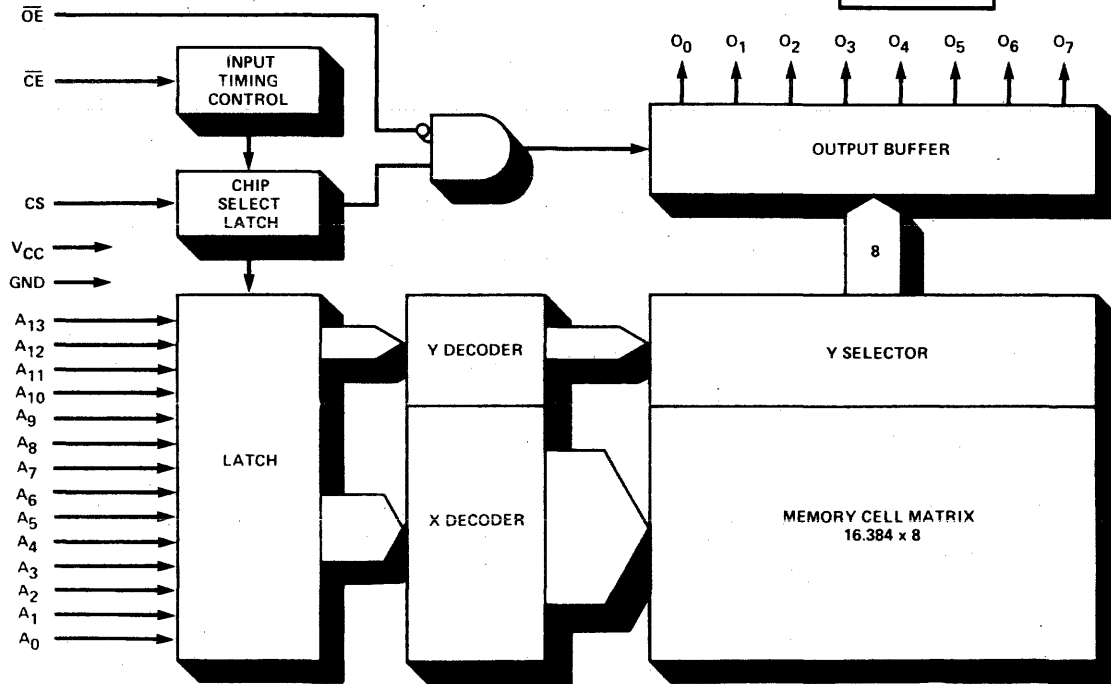
**PIN CONFIGURATION (TOP VIEW)**



**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
μPD 23C128PC	5μsec	28 Pin Molded DIP

**BLOCK DIAGRAM**



**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Ratings	Unit
$V_{CC}$	Supply Voltage	-0.3 ~ +7	V
$V_I$	Input Voltage	-0.3 ~ $V_{CC} + 0.3$	V
$V_O$	Output Voltage	-0.3 ~ $V_{CC} + 0.3$	V
$T_{opt}$	Operating Temperature	-10 ~ +70	°C
$T_{stg}$	Storage Temperature	-40 ~ +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

## D.C. CHARACTERISTICS (T<sub>a</sub> = 0° to +70°C)

μPD23C128

Symbol	Parameter	Test Condition	V <sub>CC</sub> = 5V ± 10%			V <sub>CC</sub> = 3.5 to 5.5V			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
V <sub>IH</sub>	Input "High" Voltage		0.7V <sub>CC</sub>		V <sub>CC</sub>	0.7V <sub>CC</sub>		V <sub>CC</sub>	V
V <sub>IL</sub>	Input "Low" Voltage				0.7			0.5	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -1.0mA	2.4						V
		I <sub>OH</sub> = -0.5mA	3.5						V
		I <sub>OH</sub> = -200μA				3.0			V
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = 1.0mA			0.45				V
		I <sub>OL</sub> = 200μA						0.5	V
I <sub>LIH</sub>	Input Leakage Current High	V <sub>I</sub> = V <sub>CC</sub>			1.0			1.0	μA
I <sub>LIL</sub>	Input Leakage Current Low	V <sub>I</sub> = 0V			-1.0			-1.0	μA
I <sub>LOH</sub>	Output Leakage Current High	V <sub>O</sub> = V <sub>CC</sub>			1.0			1.0	μA
I <sub>LOL</sub>	Output Leakage Current Low	V <sub>O</sub> = 0V			-1.0			-1.0	μA
I <sub>CC1</sub>	Supply Current			1	4		1	4	mA
I <sub>CC2</sub>		Chip		0.1	0.4		0.1	0.4	mA
I <sub>CC3</sub>		Deselect	*1			2			2

\*1 = No transition on input signals

## A.C. CHARACTERISTICS (T<sub>a</sub> = 0° to +70°C)

Symbol	Parameter	Condition	V <sub>C</sub> = 5V ± 10%			V <sub>CC</sub> = 3.5 to 5.5V			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
t <sub>CYC</sub>	Cycle Time	C <sub>L</sub> = 50 pF	5.5			11.0			μs
t <sub>AS</sub>	Address Setup Time		0			0			μs
t <sub>AH</sub>	Address Hold Time		1.0			2.0			μs
t <sub>ACC</sub>	Access Time				5.0			10.0	μs
t <sub>CE</sub>	Chip Enable Access Time				0.75			1.5	μs
t <sub>DF</sub>	Output Disable Time				0.5			1.0	μs
t <sub>CO</sub>	Chip Enable Off Time		0.5			1.0			μs

Symbol	Parameter	Test Condition	Limits			Unit
			Min.	Typ.	Max.	
C <sub>I</sub>	Input Capacitance	V = 0V, f = 1MHz			15	pF
C <sub>O</sub>	Output Capacitance	T <sub>a</sub> = 25°C			15	pF

## CUSTOM PROGRAMMING INSTRUCTIONS

### BIT PATTERN SUBMITTAL OPTIONS:

The customer's unique bit pattern can be submitted in several convenient methods that are easy for the ROM customer, and readily verifiable for accuracy. The bit pattern can be delivered to NEC contained within:

1. One programmed 27128 EPROM
2. Two programmed 2764 EPROMs
3. Four programmed 2732 EPROMs

### BIT PATTERN VERIFICATION:

For customer verification of the submitted bit patterns, several alternatives are also available. The following are those found to be most expeditious.

CUSTOM PATTERN SUBMITTED VIA:	VERIFICATION ROUTINE
1. One programmed 27128 EPROM	Customer sends NEC one additional erased 27128. NEC programs the spare 27128 with the data from the programmed 27128 and returns to customer for verification.
2. Two programmed 2764 EPROMs	Customer sends NEC two additional erased 2764's. NEC programs the spare 2764's with the data from the programmed 2764's and returns to the customer for verification.
3. Four programmed 2732 EPROMs	Customer sends NEC four additional erased 2732's. NEC programs the spare 2732's with the data from the programmed 2732's and returns to the customer for verification.

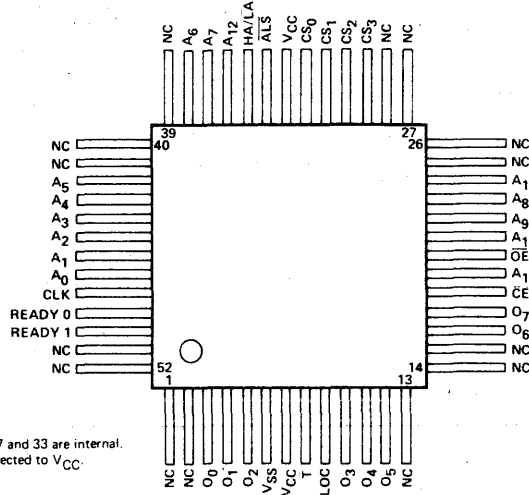
**FEATURES:**

- 16,384 Words by 8 Bits
- Access Time: 5μsec VCC " 5V ± 10%
- Low Power Consumption—CMOS Technology
- Four Mask Programmable Chip Selects
- Three State Output for Direct Bus Interface
- Built for Direct Interface with 8048/8748 μproc.
- Single Power Supply (3.5-5.5V)
- 52 Pin Plastic Flat Package

**GENERAL DESCRIPTION:**

The μPD73128G is a 131,072 bit Read Only Memory utilizing low power CMOS technology. The device is mask programmable and is organized 16,384 words, by 8 bits/word and operates from a single 3.5-5.5 volt supply. The device has four mask programmable chip select inputs for simple memory expansion and has three-state output for direct bus interface.

**PIN CONFIGURATION (TOP VIEW)**

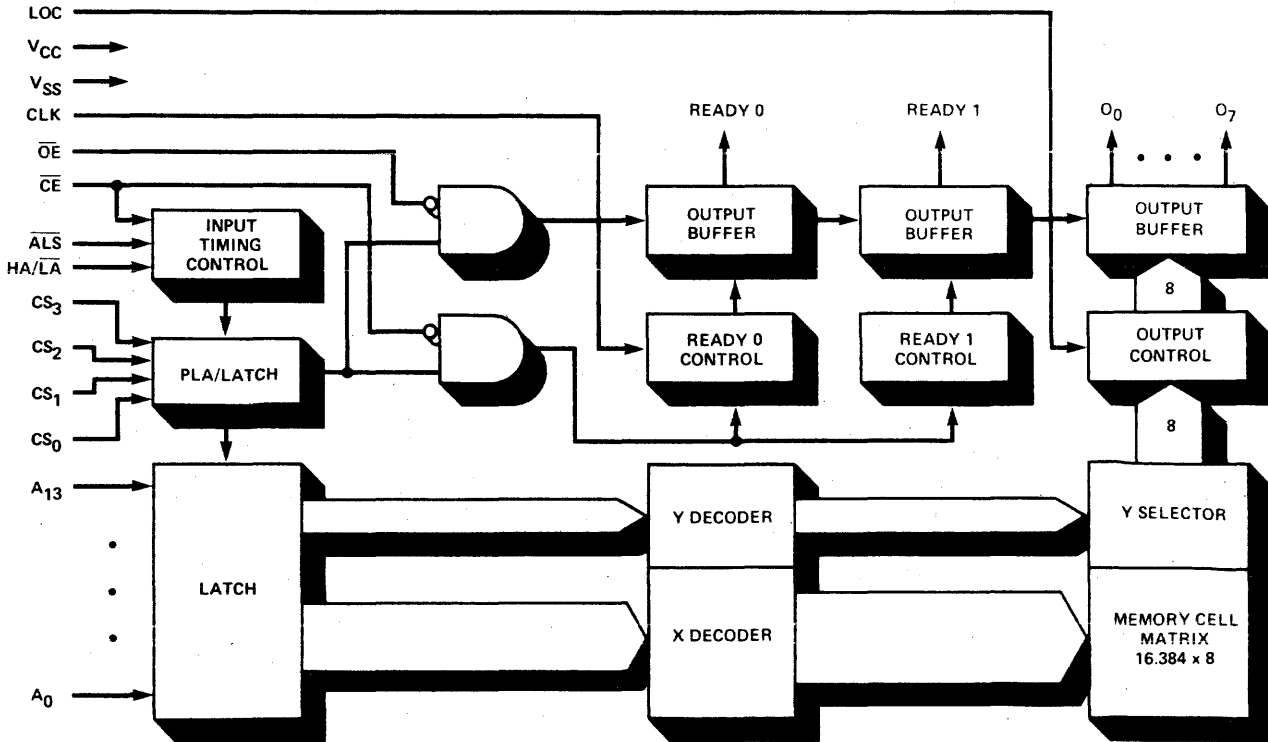


NOTE: Pins 7 and 33 are internal.  
Connected to VCC.

**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
μPD 73128G	5μsec	52 Pin Plastic Flat Package

**BLOCK DIAGRAM**



MEMORY  
NEC Electronics

# ABSOLUTE MAXIMUM RATINGS

**μPD73128G**

Symbol	Parameter	Ratings	Unit
V <sub>CC</sub>	Supply Voltage	-0.3 ~ +7	V
V <sub>I</sub>	Input Voltage	-0.3 ~ V <sub>CC</sub> +0.3	V
V <sub>O</sub>	Output Voltage	-0.3 ~ V <sub>CC</sub> +0.3	V
T <sub>opt</sub>	Operating Temperature	-10 ~ +70	°C
T <sub>stg</sub>	Storage Temperature	-40 ~ +125	°C

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

## D.C. CHARACTERISTICS (T<sub>a</sub> = 0° ~ +70° C)

Symbol	Parameter	Test Condition	V <sub>CC</sub> = 5V ± 10%			V <sub>CC</sub> = 2.5 ~ 6.0V			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
V <sub>IH</sub>	Input "High" Voltage		0.7V <sub>CC</sub>		V <sub>CC</sub>	0.7V <sub>CC</sub>		V <sub>CC</sub>	V
V <sub>IL</sub>	Input "Low" Voltage				0.8			0.4	V
V <sub>OH</sub>	Output "High" Voltage	I <sub>OH</sub> = -1.0mA	2.4						V
		I <sub>OH</sub> = -0.5mA	3.5						V
		I <sub>OH</sub> = -200μA				2.0			V
V <sub>OL</sub>	Output "Low" Voltage	I <sub>OL</sub> = 1.0mA			0.45				V
		I <sub>OL</sub> = 200μA						0.5	V
I <sub>LIH</sub>	Input Leakage Current High	V <sub>I</sub> = V <sub>CC</sub>			1.0			1.0	μA
I <sub>LIL</sub>	Input Leakage Current Low	V <sub>I</sub> = 0V			-1.0			-1.0	μA
I <sub>LOH</sub>	Output Leakage Current High	V <sub>O</sub> = V <sub>CC</sub>			1.0			1.0	μA
I <sub>LOL</sub>	Output Leakage Current Low	V <sub>O</sub> = 0V			-1.0			-1.0	μA
I <sub>CC1</sub>	Supply Current	Chip Deselected		4	10		4	12	mA
I <sub>CC2</sub>					10			10	μA

## A.C. CHARACTERISTICS (T<sub>a</sub> = 0° ~ +70° C)

Symbol	Parameter	Condition	V <sub>CC</sub> = 5V ± 10%			V <sub>CC</sub> = 2.5 ~ 6.0V			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
t <sub>C1</sub>	CE Cycle Time	C <sub>L</sub> = 50pF	*1			*1			μS
t <sub>C2</sub>			*2			*2			μS
t <sub>AS</sub>	Address Setup Time		0.5			1			μS
t <sub>AH</sub>	Address Hold Time		0.5			1			μS
t <sub>AC</sub>	Access Time				3			6	μS
t <sub>HS</sub>	ALS, HA/LA Setup Time		0.75			1.5			μS
t <sub>HH</sub>	ALS, HA/LA Hold Time		0.75			1.5			μS
t <sub>OE</sub>	Data Output Delay Time (from OE)				0.75			1.5	μS
t <sub>ON1</sub>	READY Output Delay Time (from CE)				0.75			1.5	μS
t <sub>ON2</sub>	READY Output Delay Time (from OE)				0.75			1.5	μS
t <sub>RD</sub>	READY Enable to Output Delay Time (from CE)				5			10	μS
t <sub>OF1</sub>	Output Float Delay Time (from CE)				0.5			1	μS
t <sub>OF2</sub>	Output Float Delay Time (from OE)				0.5			1	μS
t <sub>p</sub>	CE Precharge Time			1			2		μS
Symbol	Parameter	Test Condition	Limits			Unit			
			Min.	Typ.	Max.				
C <sub>I</sub>	Input Capacitance	V <sub>I</sub> = 0V, f = 1MHz			15	pF			
C <sub>O</sub>	Output Capacitance	V <sub>O</sub> = 0V, f = 1MHz			15	pF			

\*1 = t<sub>C1</sub> = t<sub>p</sub> + t<sub>AC</sub>

\*2 = t<sub>C2</sub> = t<sub>C1</sub> + t<sub>p</sub> + t<sub>HH</sub> + t<sub>HS</sub>

Additional Application Information Is Available.

**FEATURES:**

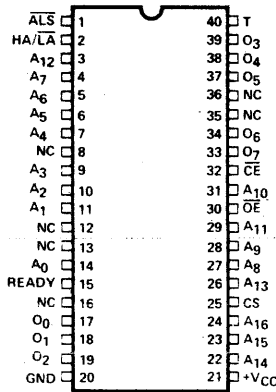
- One Megabit Programmable Data Capacity
- 131,072 Words by 8 Bits
- Access Time: 3μsec  $V_{CC} = 5V \pm 10\%$
- Low Power Consumption—CMOS Technology
- Two Mask Programmable Chip Selects
- Three State Output for Direct Bus Interface
- Built for Direct Interface with 8048/8748 μprocessor
- Single Power Supply (3.5-5.5V)
- 52 Pin Plastic Flat Package/40 Pin DIP

**GENERAL DESCRIPTION:**

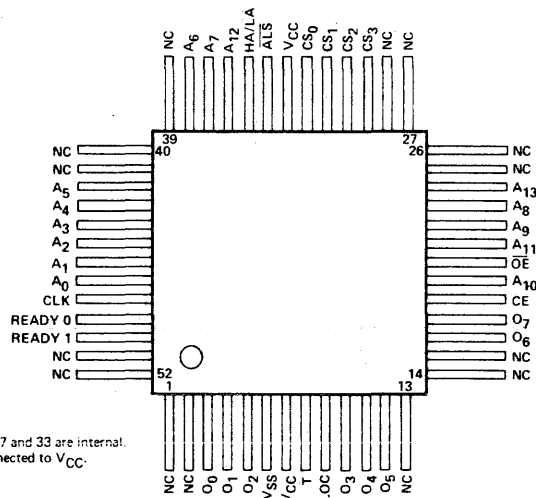
The μPD731000 is a 131,072 bit Read Only Memory utilizing low power CMOS technology. The device is mask programmable and is organized 16,384 words, by 8 bits/word and operates from a single 3.5-5.5 volt supply. The device has one mask programmable chip select inputs for simple memory expansion and has three-state outputs for direct bus interface.

**ORDERING INFORMATION**

PART NUMBER	ACCESS TIME	PACKAGE
μPD 731000G	5μsec	52 Pin Plastic Flat Package
μPD 738000PC		40 Pin Plastic DIP

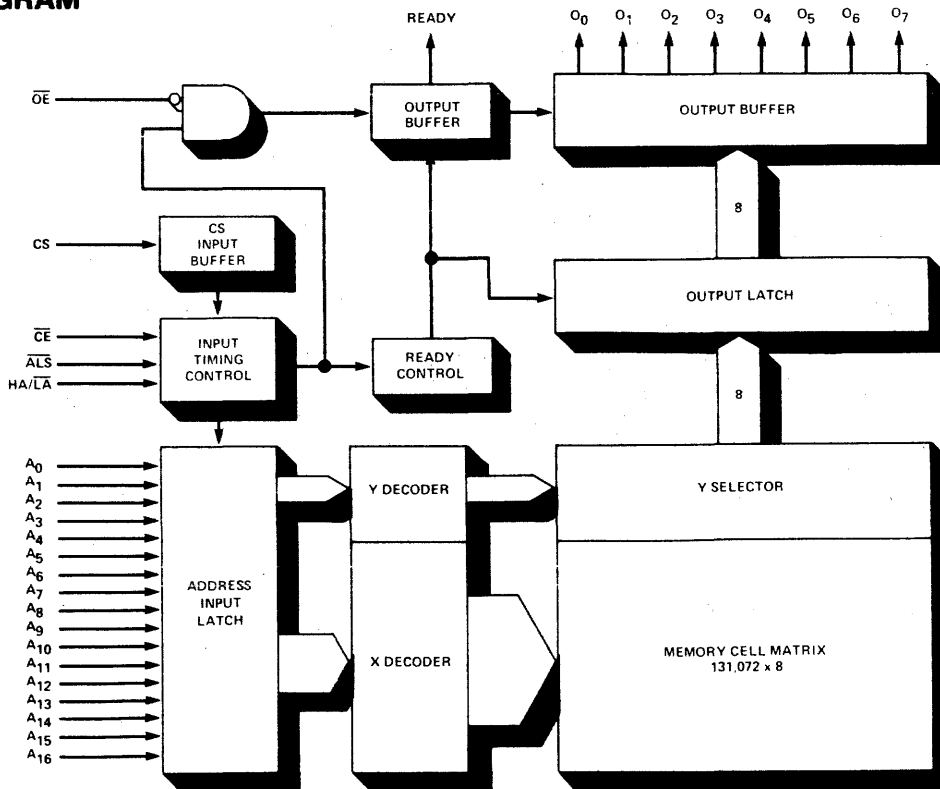


**PIN CONFIGURATION (TOP VIEW)**



NOTE: Pins 7 and 33 are internal. Connected to  $V_{CC}$ .

**BLOCK DIAGRAM**



MEMORY NEC Electronics



**ABSOLUTE MAXIMUM RATINGS** ( $T_a = 25^\circ\text{C}$ )

**$\mu\text{PD731000}$**

Symbol	Parameter	Ratings	Unit
$V_{CC}$	Supply Voltage	-0.5 to +7	V
$V_i$	Input Voltage	-0.5 to +7	V
$V_o$	Output Voltage	-0.5 to +7	V
$T_{opt}$	Operating Temperature	-0 to +70	$^\circ\text{C}$
$T_{stg}$	Storage Temperature	-65 to +125	$^\circ\text{C}$

Stresses more severe than those listed here may cause permanent damage to the device. This is a stress rating only, and operation of the device at any condition above those indicated in the operational sections of this specification is not implied.

**D.C. CHARACTERISTICS** ( $T_a = 0^\circ \sim +70^\circ\text{C}$ ,  $V_{SS} = 0\text{V}$ )

Symbol	Parameter	Test Condition	$V_{CC} = 5\text{V} \pm 10\%$			$V_{CC} = 2.7 \sim 5.5\text{V}$			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
$V_{IH}$	Input "High" Voltage		0.7 $V_{CC}$		$V_{CC}$	0.7 $V_{CC}$		$V_{CC}$	V
$V_{IL}$	Input "Low" Voltage				0.8			0.4	V
$V_{OH}$	Output "High" Voltage	$I_{OH} = -1.0\text{mA}$	3.5						V
									V
$V_{OL}$	Output "Low" Voltage	$I_{OH} = -200\mu\text{A}$				2.0			V
									V
$V_{OL}$	Output "Low" Voltage	$I_{OL} = 1.0\text{mA}$			0.4				V
		$I_{OL} = 200\mu\text{A}$						0.5	V
$I_{LIH}$	Input Leakage Current High	$V_i = V_{CC}$			1.0			1.0	$\mu\text{A}$
$I_{LIL}$	Input Leakage Current Low	$V_i = 0\text{V}$			-1.0			-1.0	$\mu\text{A}$
$I_{LOH}$	Output Leakage Current High	$V_o = V_{CC}$			1.0			1.0	$\mu\text{A}$
$I_{LOL}$	Output Leakage Current Low	$V_o = 0\text{V}$			-1.0			-1.0	$\mu\text{A}$
$I_{CC1}$	Supply Current			5	10		5	12	$\text{mA}$
$I_{CC2}$		*1	*2	0.1	10		0.1	10	$\mu\text{A}$
$I_{CC3}$					10			10	$\mu\text{A}$
Symbol	Parameter	Test Condition	Limits			Unit			
			Min.	Typ.	Max.				
$C_i$	Input Capacitance	$V_i = 0\text{V}$ , $f = 1\text{MHz}$				15	$\text{pF}$		
$C_o$	Output Capacitance	$V_o = 0\text{V}$ , $f = 1\text{MHz}$				15	$\text{pF}$		

- \*1 = Chip is deselected
- \*2 = Input signal is not changed

**A.C. CHARACTERISTICS** ( $T_a = 0^\circ \sim +70^\circ\text{C}$ ,  $V_{SS} = 0\text{V}$ )

Symbol	Parameter	Condition	$V_{CC} = 5\text{V} \pm 10\%$			$V_{CC} = 2.5 \sim 6.0\text{V}$			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
$t_{C1}$	CE Cycle Time	$C_L = 50\text{pF}$	*3			*3			$\mu\text{s}$
$t_{C2}$			*4			*4			$\mu\text{s}$
$t_{AS}$	Address Setup Time		0.5			1			$\mu\text{s}$
$t_{AH}$	Address Hold Time		0.5			1			$\mu\text{s}$
$t_{ACC}$	Access Time				3			6	$\mu\text{s}$
$t_{HS}$	ALS, HA/LA Setup Time		0.75			1.5			$\mu\text{s}$
$t_{HH}$	ALS, HA/LA Hold Time		0.75			1.5			$\mu\text{s}$
$t_{CSS}$	CS Setup Time		1			2			$\mu\text{s}$
$t_{OE}$	Output Enable Access Time				0.75			1.5	$\mu\text{s}$
$t_{ON1}$	READY Output Delay Time (from CE)				0.75			1.5	$\mu\text{s}$
$t_{ON2}$	READY Output Delay Time (from OE)			0.75			1.5	$\mu\text{s}$	
$t_{RD}$	READY Enable to Output Delay Time (from CE)			5			10	$\mu\text{s}$	
$t_{OF1}$	Output Float Delay Time (from CE)			0.5			1	$\mu\text{s}$	
$t_{OF2}$	Output Float Delay Time (from OE)			0.5			1	$\mu\text{s}$	
$t_p$	CE Precharge Time		1			2		$\mu\text{s}$	

\*3 =  $t_{C1} = t_p + t_{AC}$   
 \*4 =  $t_{C2} = t_{C1} + t_p + t_{HH} + t_{HS}$   
 Additional application information is available.

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PART NUMBER	DESCRIPTION
<b>CMOS MICROPROCESSOR</b>	
MSM80C85A RS	8 BIT CMOS MICROPROCESSOR
MSM80C85A GSK	8 BIT CMOS MICROPROCESSOR
MSM80C86 RS	16 BIT CMOS MICROPROCESSOR
MSM80C86-2 RS	16 BIT CMOS MICROPROCESSOR
MSM80C87 RS	CMOS NUM. DATA PROC.
MSM80C88 RS	8 BIT CMOS MICROPROCESSOR
MSM80C88-2 RS	8 BIT CMOS MICROPROCESSOR
MSM80C188 RS	8 BIT CMOS MICROPROCESSOR

PART NUMBER	DESCRIPTION
<b>NMOS MICROPROCESSOR</b>	
MSM8085A RS	8 BIT NMOS MICROPROCESSOR

PART NUMBER	DESCRIPTION
<b>CMOS MICROCOMPUTERS</b>	
MSM5840H RS	4 BIT CMOS MCU—NO ROM
MSM5840H-XX RS	4 BIT CMOS MCU—W/ROM
MSM5840H-XX GS	4 BIT CMOS MCU—W/ROM
MSM5842-XX RS	4 BIT CMOS MCU—W/ROM
MSM5842-XX GSK	4 BIT CMOS MCU—W/ROM
MSM58421-XX GSK	4 BIT CMOS MCU-LCD—W/ROM
MSM58422-XX GSK	4 BIT CMOS MCU-VF—W/ROM
MSM5845-XX RS	4 BIT CMOS MCU—W/ROM
MSM80C35 RS	ROMLESS CMOS 80C48
MSM80C39 RS	ROMLESS CMOS 80C49
MSM80C48-XX RS	CMOS 8048 8 BIT UC
MSM80C48-XX GSK	CMOS 8048 8 BIT UC
MSM80C49-XX RS	CMOS 8049 8 BIT UC
MSM80C49-XX GSK	CMOS 8049 8 BIT UC

PART NUMBER	DESCRIPTION
<b>MCU/MPU DEVELOPMENT PACKAGES</b>	
EASE-40-I	SERIES 40 DEV PKG—ISIS
EASE-40-C	SERIES 40 DEV PKG—CP/M
EASE-49-I	SERIES 80 DEV PKG—ISIS
EASE-49-C	SERIES 80 DEV PKG—CP/M
MPB-202	SERIES 40 DEV BOARD
MPB-203	SERIES 40 DEV BOARD
MPB-421-LCD	SERIES 40 DEV BOARD
MPB-422-VF	SERIES 40 DEV BOARD
MPB-802	SERIES 80 DEV BOARD

PART NUMBER	DESCRIPTION
<b>PERIPHERAL CIRCUITS</b>	
MSM3914A-34 RS	NMOS 16X8 KYBD ENCDR
MSM5204A RS	8 BIT CMOS A/D CONV
MSM5204-1 KIT	5204/450 KHZ RESONATOR
MSM5832 RS	CMOS REAL TIME CLOCK
MSM58321 RS	CMOS REAL TIME CLOCK
MSM6203 RS	ARITHMETIC CIRCUIT
MSM80C58 RS	CMOS ARITHMETIC CKT
MSM8155 RS	NMOS 256X8 RAM/I-O TIMR
MSM81C55 RS	CMOS 256X8 RAM/I-O/TIMR
MSM82C12 RS	CMOS 8 BIT I/O PORT
MSM82C43 RS	CMOS 8 BIT I/O PORT EXPDR
MSM82C51A RS	CMOS USART
MSM82C53-5 RS	CMOS PROG. INTVR TIMER
MSM82C55A-5 RS	CMOS PROG. PERIP INTFCE
MSM8255 RS	256X8 RAM/I-O/TIMER
MSM82C57-5 RS	CMOS DMA CONTROLLER
MSM82C59A-2 RS	CMOS INTERRUPT CONTROLLER
MSM82C84A RS	CMOS CLOCK GEN DEV
MSM82C88 RS	CMOS BUS CNTL
MSM83C55-XX RS	CMOS 16K ROM, I/O
MSM83C55-2 RS	CMOS 16K ROM + I/O

PART NUMBER	DESCRIPTION
<b>DYNAMIC RAM</b>	
MSM3732L-15 RS	32KX1 NMOS DRAM
MSM3732H-15 RS	32KX1 NMOS DRAM
MSM3732L-20 RS	32KX1 NMOS DRAM
MSM3732H-20 RS	32KX1 NMOS DRAM
MSM3764-12 RS	64KX1 NMOS DRAM
MSM3764-15 RS	64KX1 NMOS DRAM
MSM3764-20 RS	64KX1 NMOS DRAM
MSM41256 RS	256KX1 NMOS DRAM
MSM37256 AS	256KX1 NMOS DRAM
MSM41257 RS	256KX1 NMOS DRAM

PART NUMBER	DESCRIPTION
<b>STATIC RAM</b>	
MSM5114-2 RS	1KX4 CMOS STATIC RAM
MSM5128-12 RS	2KX8 CMOS STATIC RAM
MSM5128-15 RS	2KX8 CMOS STATIC RAM
MSM5128-15 GSK	2KX8 CMOS STATIC RAM
MSM5128-20 RS	2KX8 CMOS STATIC RAM
MSM5128-20 GSK	2KX8 CMOS STATIC RAM
MSM5165 RS	8KX8 CMOS STATIC RAM

PART NUMBER	DESCRIPTION
<b>EPROM</b>	
MSM2764 AS	8KX8 NMOS EPROM
MSM27C64 AS	8KX8 CMOS EPROM

PART NUMBER	DESCRIPTION
<b>ROM</b>	
MSM38256 RS	32KX8 NMOS MASK ROM
MSM3864 RS	8KX8 NMOS MASK ROM

PART NUMBER	DESCRIPTION
<b>CMOS VOICE SYNTHESIZERS</b>	
MSM5205 RS	SPEECH SYNTHESIZER, ADPCM
MSM5205-2 KIT	5205/ALP-2 FILT/RESONTR
MSM5205-3 KIT	5025/ALP-3 FILT/RESONTR
MSM5205-4 KIT	5205/ALP-4 FILT/RESONTR
MSM5218 RS	SPEECH ANALY/SYNTHESIS, ADPCM
MSM5218-2 KIT	5218/ALP-2 FILT/RESONTR
MSM5218-2-2 KIT	5218/TWO ALP-2 FIL/RES 1
MSM5218-3 KIT	5218/ALP-3 FILT/RESONTR
MSM5218-3-3 KIT	5218/TWO ALP-3 FIL/RES 1
MSM218-4 KIT	5218/ALP-4 FILT/RESONTR
MSM5218-4-4 KIT	5218/TWO ALP-4 FIL/RES 1
MSM5204A RS	8 BIT CMOS A/D CONVERTER
MSM5248 RS	SPEECH SYNTH, 48K BIT ROM, ADPCM
MSM6202 GSK	SPEECH SYNTH, 144K BIT ROM, ADPCM
MSM6212 RS	SPEECH SYNTH, 288K BIT ROM, ADPCM
MSM6243 GSK	SPEECH SYNTH, 192K BIT ROM, ADPCM
MSM6912 AS	AUDIO FILTER
SAS-1	SPEECH DEV SYSTEM
SAS-2	SPEECH DEV SYSTEM
ALP-2	LOW PASS FILTER 2KHz
ALP-3	LOW PASS FILTER 3KHz
ALP-4	LOW PASS FILTER 4KHz

PART NUMBER	DESCRIPTION
<b>TELECOMMUNICATIONS</b>	
MSA104 RS	LOGARITHMIC AMPLIFIER
MSA105 AS	LOGARITHMIC AMPLIFIER
MSM5249 RS	CMOS PULSE DIALER
MSM6910 AS	CMOS CODEC/FILT COMBO
MSM6912 AS	PCM CHANNEL FILTER
MSM6912 RS	PCM CHANNEL AUDIO FILT
MSM6917 AS	U-255 CODEC CIRCUIT
MSM3953 AS	1200B CCITT MODEM
MSM6926 RS	300B CCITT MODEM
MSM6927 RS	1200B CCITT MODEM
MSM6946 RS	300B BELL MODEM
MSM6947 RS	1200B BELL MODEM

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MEMORY

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PART NUMBER	DESCRIPTION
<b>DISPLAY DRIVERS &amp; CHARACTER GENERATORS</b>	
MSL912	RS OCTAL +30V DISP DRVR
MSL915	RS OCTAL -60V DISP DRVR
MSL915	DICE OCTAL -60V DISP DRVR
MSL917	RS OCTAL -60V DISP DRVR
MSL966	RS LED DIGIT DRIVER
MSM5219	GSK CMOS 48 BIT LCD DRIVER
MSM5219	DICE CMOS 48 BIT LCD DRIVER
MSM5238	GSK 32 ROW LCD D/MATRIX DR
MSM5526	RS RF COUNTER/DISP DRIVER
MSM58282	RS 4D LED DISPLAY DRIVER
MSM58283	RS 4D VF DISPLAY DRIVER
MSM58292	GS 5D LCD DISPLAY DRIVER
MSM58293	GS 5D VF DISPLAY DRIVER
MSM58371	RS 12 BIT S/R W/LED DRV
MSM5838	GS LCD DOT ROW DRIVER
MSM5839	GS LCD DOT COL DRIVER
MSM9353	RS LCD 8 DOT LEVEL METER
MSM9356	GSK LCD 32 DOT LEVEL METER
MSM9356	DICE LCD 32 DOT LEVEL METER
MSL9510	RS 8D SCAN VF DISP DRVR
MSL9511	RS 8D SCAN VF DISP DRVR
MSL9650-01	AS 35 DOT CHAR GEN + ROM
MSL9662-01	RS 16 SEG AN CHAR GEN
MSL9663-01	RS 16 SEG AN CHAR GEN
MSL9664-01	RS 14 SEG AN CHAR GEN
MSL9665-01	RS 14 SEG AN CHAR GEN

PART NUMBER	DESCRIPTION
<b>TIMEKEEPING</b>	
MSM5509	RS 6D LED DIGITAL CLOCK
MSM5523	RS LED CLK, RF CNTR, 4.5 DIG
MSM5524	RS 4 1/2 DIG VF CLOCK
MSM5525	RS 3 1/2 DIG VF RF CNTR
MSM5528	RS 3 1/2 DIG VF CAR CLOCK
MSM5529	RS 4D MUX VF AUTO CLOCK
MSM5550	RS 4D VF ALARM CLK CKT
MSM5557	GSK MUX LCD ALARM CLOCK
MSM5557	DICE MUX LCD ALARM CLOCK
MSM5558	RS VF DIGITAL CLOCK
MSM5803	GSK LCD ALARM CLOCK

<b>MISCELLANEOUS CIRCUITS</b>	
MSL2312	RS DIV 10/100 PRESCALER
MSM5232	RS 8 CHANNEL TONE GEN
MSM5814	RS MOTOR CONTROLLER
MSM5816	RS MOTOR CONTROL PLL
MSM5819	RS MOTOR CONTROL PLL
MSL9362	RS 4 CH PROP R/C XMTR
MSL9363	RS 4 CH PROP R/C RCVR

PACKAGING RS: PLASTIC GS: PLASTIC FLAT PACK  
AS: CERAMIC GSK: BENT LEAD FLAT PACK

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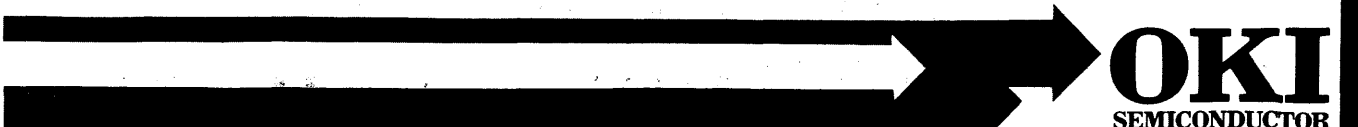
**OKI Semiconductor, Inc.**  
65 Williams Street, Suite 120  
Wellesley, Massachusetts 02181  
Tel: 617/431-7944 TWX: 710/383-6812

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**OKI Semiconductor, Inc.**  
205 South Avenue, Suite 100  
Poughkeepsie, New York 12601  
Tel: 914/473-8044 TLX: 0291516

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OKI Semiconductor MEMORY

## MN4164, N MOS 65K Bit Dynamic Ram

### Description

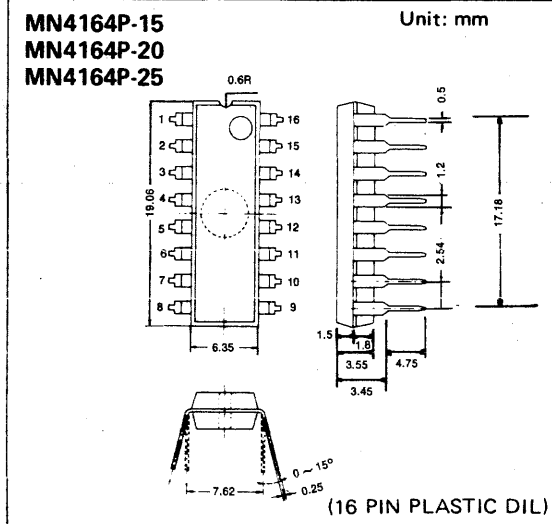
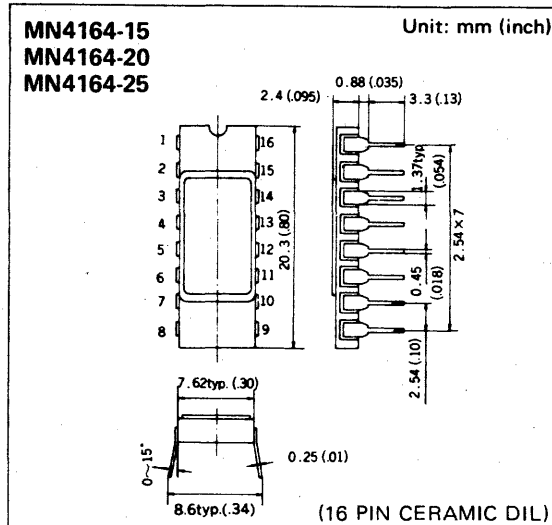
System oriented features include: operation from a single +5V  $\pm 10\%$  tolerance power supply, direct TTL interfacing capability, on-chip addresses and data registers which eliminate the need for interface registers, and two chip select methods to allow the user to determine the appropriate speed/power characteristics of his memory system.

The RAM module also incorporates several flexible operating modes: "Read," "Write," "Read-Modify-Write" cycles, "Page-Mode" operation and "RAS-Only" refresh. Proper control of the clock inputs (RAS, CAS and WRITE) allows common I/O capability, two dimensional chip selection, and extended page boundaries (for operating in page mode).

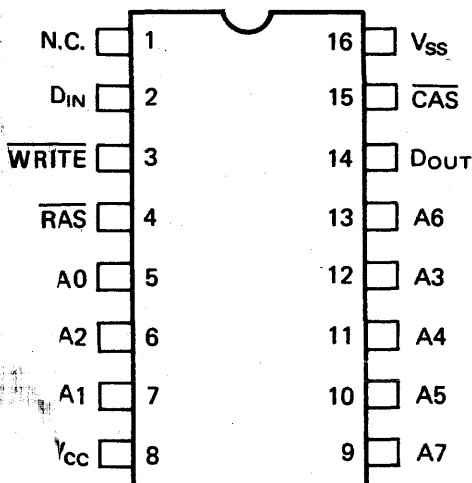
### Features

- 65,536 x 1 RAM, 16-pin package
- Row access time:
  - 150 ns Max. (MN4164-15/MN4164P-15)
  - 200 ns Max. (MN4164-20/MN4164P-20)
  - 250 ns Max. (MN4164-25/MN4164P-25)
- Cycle time:
  - 270 ns Max. (MN4164-15/MN4164P-15)
  - 330 ns Max. (MN4164-20/MN4164P-20)
  - 410 ns Max. (MN4164-25/MN4164P-25)
- Low power dissipation:
  - 275 mW Max. (active)
  - 27.5 mW Max. (standby)
- Single 5V supply,  $\pm 10\%$  tolerance
- 128 refresh cycles/2ms

### MN4164 Block Diagram



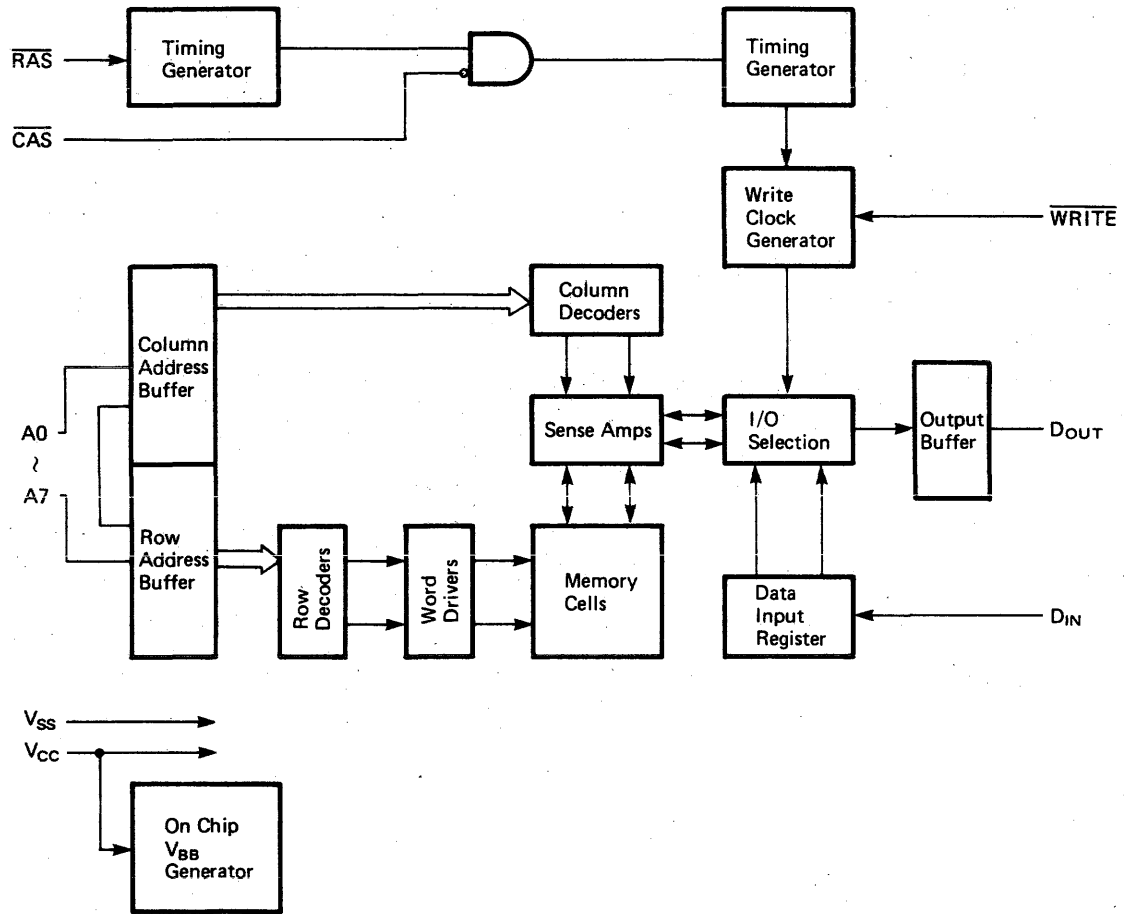
### Pin Assignment (Top View)



Pin Names	Function
A <sub>0</sub> ~ A <sub>7</sub>	Address Inputs
RAS	Row Address Strobe
CAS	Column Address Strobe
WRITE	Write Enable
D <sub>IN</sub>	Data Input
D <sub>OUT</sub>	Data Output
V <sub>CC</sub>	Power (+5V)
V <sub>SS</sub>	Ground (0V)

The above specifications are subject to change without prior notice. While every precaution has been taken in the preparation of this data sheet, the publisher assumes no responsibility for patent liability with respect to the use of the information contained herein.

## MN4164, N MOS 65K Bit Dynamic Ram



### Absolute Maximum Ratings

Rating	Symbol	Value	Unit
Voltage on any Pin relative to $V_{SS}$	$V_{IN}, V_{OUT}$	-1.0 to +7.0	V
Voltage on $V_{CC}$ Supply relative to $V_{SS}$	$V_{CC}$	-1.0 to +7.0	V
Operating Temperature	$T_{OP}$	0 to +70	°C
Storage Temperature	$T_{STG}$	-55 to +150	°C
Power Dissipation	$P_D$	1	W
Short Circuit Current	$I_{OS}$	50	mA

Note: Exceeding Absolute Maximum Ratings may cause permanent device damage. Functional operating of the device is not implied outside the operating conditions. Exposure to absolute maximum ratings for extended periods of time may impact device reliability.

### Recommended Operating Conditions (Referenced to $V_{SS}$ )

Parameter	Symbol	Min	Typ	Max	Unit	Temperature
Supply Voltage	$V_{CC}$	4.5	5.0	5.5	V	0°C to +70°C
	$V_{SS}$	0	0	0	V	
Input High Voltage, all inputs	$V_{IH}$	2.2	—	$V_{CC} + 1.0V$	V	
Input Low Voltage, all inputs	$V_{IL}$	-1.0	—	0.8	V	

### Capacitance ( $T_a = 25^\circ\text{C}, f = 1\text{ MHz}$ )

Parameter	Symbol	Min	Typ	Max	Unit
Input Capacitance	$C_{IN}$	—	—	10	pF
Output Capacitance	$C_{OUT}$	—	—	12	pF

## MN4164, N MOS 65K Bit Dynamic Ram

### DC Characteristics (Recommended operating conditions unless otherwise noted.)

Parameter	Symbol	Min	Typ	Max	Unit	Note
Operating Current: Average power supply current (RAS, CAS cycling; $t_{RC} = \text{min}$ )	$I_{CC1}$	—	—	50	mA	1)
Standby Current: Power supply current (RAS = CAS = $V_{IH}$ )	$I_{CC2}$	—	—	5	mA	
Refresh Current: Average supply current (RAS cycling, CAS = $V_{IH}$ ; $t_{RC} = \text{min}$ )	$I_{CC3}$	—	—	40	mA	1)
Page Mode Current: Average power supply current (RAS = $V_{IL}$ , CAS cycling; $t_{PC} = \text{min}$ )	$I_{CC4}$	—	—	40	mA	1)
Input Leakage Current: Input leakage current, any input ( $0V \leq V_{IN} \leq 5.5V$ , all other pins not under test = $0V$ )	$I_{LI}$	-10	0.1	10	$\mu A$	
Output Leakage Current: (Data out is disable, $0V \leq V_{OUT} \leq 5.5V$ )	$I_{LO}$	-10	0.1	10	$\mu A$	
Output Level: Output low voltage ( $I_{OL} = 4.2 \text{ mA}$ )	$V_{OL}$	—	—	0.4	V	
Output Level: Output high voltage ( $I_{OH} = -5 \text{ mA}$ )	$V_{OH}$	2.4	—	—	V	

Note: 1)  $I_{CC1}$ ,  $I_{CC3}$ ,  $I_{CC4}$  depend on cycle rate and output loading. Specifications are for maximum cycle rate and no load. Supply current may be scaled according to the following equation:

$$I(t_{RC}) = \frac{t_{RC \text{ min}} \times I(t_{RC \text{ min}}) + (t_{RC} - t_{RC \text{ min}}) \times I_{CC2}}{t_{RC}}$$

### AC Characteristics<sup>1), 2)</sup> (Recommended operating conditions unless otherwise noted.)

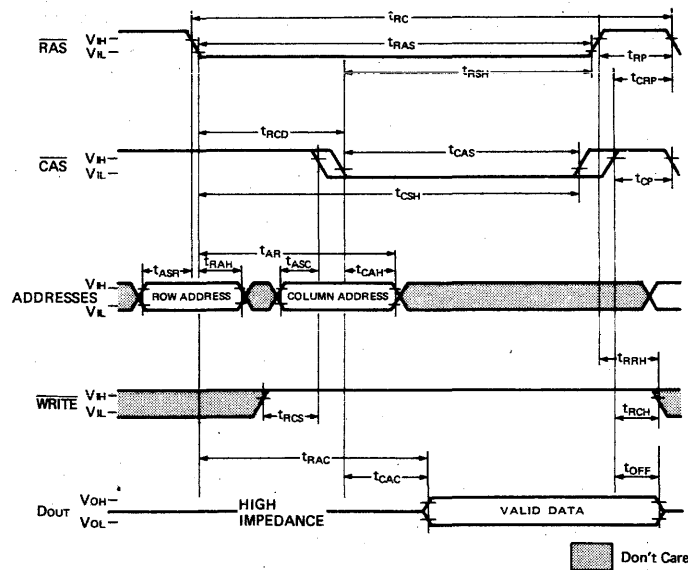
Parameter	Symbol	Unit	MN4164-15 MN4164P-15		MN4164-20 MN4164P-20		MN4164-25 MN4164P-25		Notes
			Min	Max	Min	Max	Min	Max	
Refresh period	$t_{REF}$	ms	—	2	—	2	—	2	
Random read or write cycle time	$t_{RC}$	ns	270	—	330	—	410	—	
Read-write cycle time	$t_{RWC}$	ns	310	—	375	—	515	—	
Page mode cycle time	$t_{PC}$	ns	170	—	225	—	275	—	
Access time from RAS	$t_{RAC}$	ns	—	150	—	200	—	250	4), 6), 8)
Access time from CAS	$t_{CAC}$	ns	—	100	—	135	—	165	5), 6), 8)
Output buffer turn-off delay	$t_{OFF}$	ns	0	40	0	50	0	60	7)
Transition time	$t_T$	ns	3	50	3	50	3	50	3)
RAS precharge time	$t_{RP}$	ns	100	—	120	—	150	—	
RAS pulse width	$t_{RAS}$	ns	150	10,000	200	10,000	250	10,000	
RAS hold time	$t_{RSH}$	ns	100	—	135	—	165	—	
CAS precharge time	$t_{CP}$	ns	50	—	80	—	100	—	
CAS pulse width	$t_{CAS}$	ns	100	10,000	135	10,000	165	10,000	
CAS hold time	$t_{CSH}$	ns	150	—	200	—	250	—	
RAS to CAS delay time	$t_{RCD}$	ns	25	50	25	65	40	85	8)
CAS to RAS precharge time	$t_{CRP}$	ns	-20	—	-20	—	-20	—	
Row Address set-up time	$t_{ASR}$	ns	0	—	0	—	0	—	
Row Address hold time	$t_{RAH}$	ns	20	—	20	—	35	—	
Column Address set-up time	$t_{ASC}$	ns	-5	—	-5	—	-5	—	
Column Address hold time	$t_{CAH}$	ns	45	—	55	—	75	—	
Column Address hold time referenced to RAS	$t_{AR}$	ns	95	—	120	—	160	—	
Read command setup time	$t_{RCS}$	ns	0	—	0	—	0	—	
Read command hold time	$t_{RCH}$	ns	0	—	0	—	0	—	
Write command set-up time	$t_{WCS}$	ns	-20	—	-20	—	-20	—	10)
Write command hold time	$t_{WCH}$	ns	45	—	55	—	75	—	
Write command hold time referenced to RAS	$t_{WCR}$	ns	95	—	120	—	160	—	
Write command pulse width	$t_{WP}$	ns	45	—	55	—	75	—	
Write command to RAS lead time	$t_{RWL}$	ns	60	—	80	—	100	—	
Write command to CAS lead time	$t_{CWL}$	ns	60	—	80	—	100	—	
Data-in set-up time	$t_{DS}$	ns	0	—	0	—	0	—	9)
Data-in hold time	$t_{DH}$	ns	45	—	55	—	75	—	9)
Data-in hold time referenced to RAS	$t_{DHR}$	ns	95	—	120	—	160	—	
CAS to WRITE delay	$t_{CWD}$	ns	80	—	95	—	125	—	10)
RAS to WRITE delay	$t_{RWD}$	ns	130	—	160	—	200	—	10)
Read command hold time referenced to RAS	$t_{RRH}$	ns	20	—	25	—	35	—	

## MN4164, N MOS 65K Bit Dynamic Ram

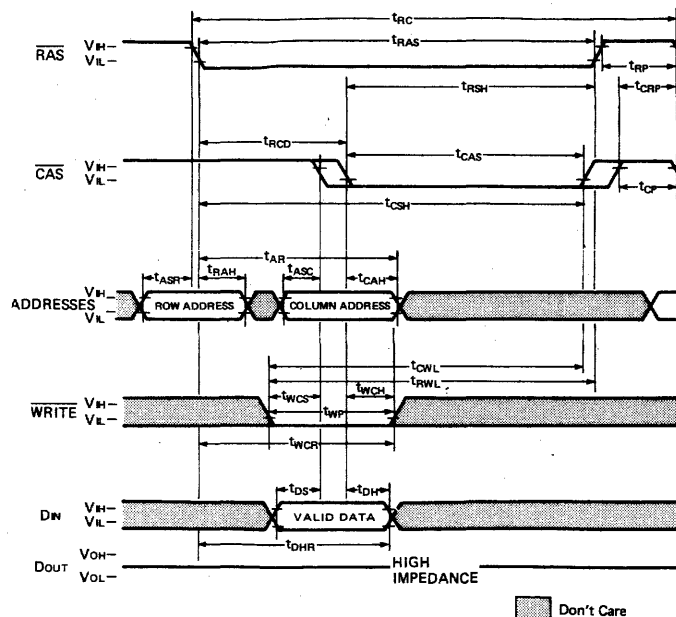
### Note:

- Several cycles are required after power up or prolonged periods of RAS inactivity (2 ms) before proper device operation is achieved. Any 8 cycles which perform refresh are adequate for this purpose.
- AC measurements assume  $t_f = 5.0$  ns.
- $V_{IH}$  and  $V_{IL}$  are reference levels for measuring signal timings and transition times.
- Assumes that  $t_{RCD} \leq t_{RCD}(\text{Max})$ .
- Assumes that  $t_{RCD} \geq t_{RCD}(\text{Max})$ .
- Measure with load equivalent to 2 TTL loads and 100 pF.
- $t_{OFF}$  defines the time at which the output enters high impedance state. It is not referenced to levels of  $V_{IH}$  and  $V_{IL}$ .
- Operation within the  $t_{RCD}(\text{Max})$  limit ensures that  $t_{RAC}(\text{Max})$  can be met.  $t_{RCD}$  is specified as a reference point only; if  $t_{RCD}$  is greater than the specified  $t_{RCD}$  limit, then row access time is  $t_{RCD} + t_{CAC}$ .
- These parameters are referenced to leading edge of CAS (Early-Write) or WRITE (Delayed-Write or Read-Modify-Write) whichever occurs last.
- $t_{WCS}$ ,  $t_{CWD}$  and  $t_{RWD}$  are not restrictive operating parameters. They are included under AC Characteristics as electrical characteristics only. If  $t_{WCS} \leq t_{WCS}(\text{Min})$ , the cycle is an Early-Write cycle and the data out pins will remain open circuit (high impedance) throughout the entire cycle. If  $t_{CWD} \geq t_{CWD}(\text{Min})$  and  $t_{RWD} \geq t_{RWD}(\text{Min})$ , the cycle is a Read-Modify-Write cycle and the data out will contain data read from the selected cell. If neither of the above sets of conditions is satisfied, the condition of the data out (at access time) is indeterminate.

### Read Cycle

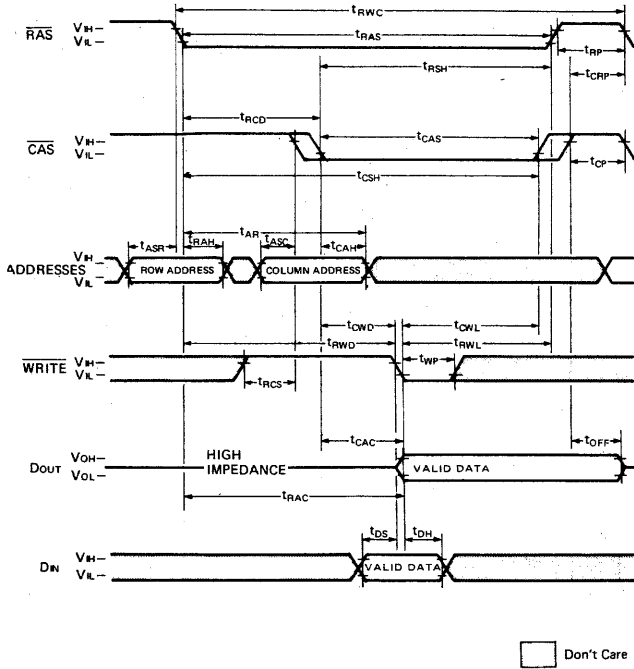


### Write Cycle (Early Write)

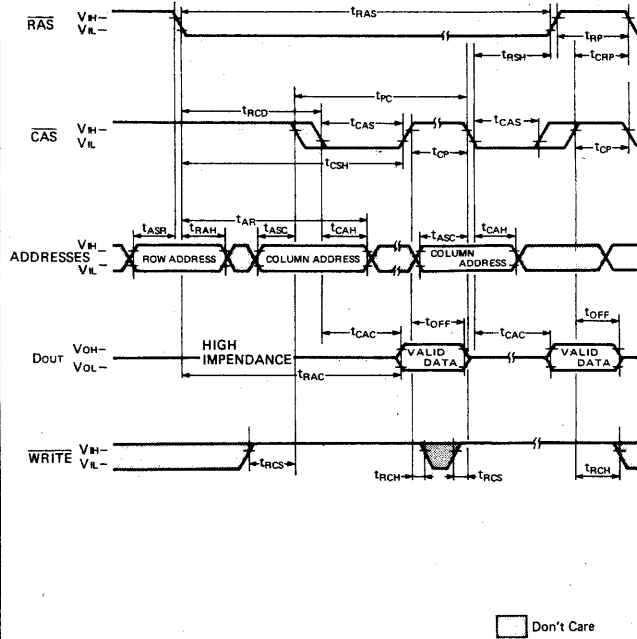


## MN4164, N MOS 65K Bit Dynamic Ram

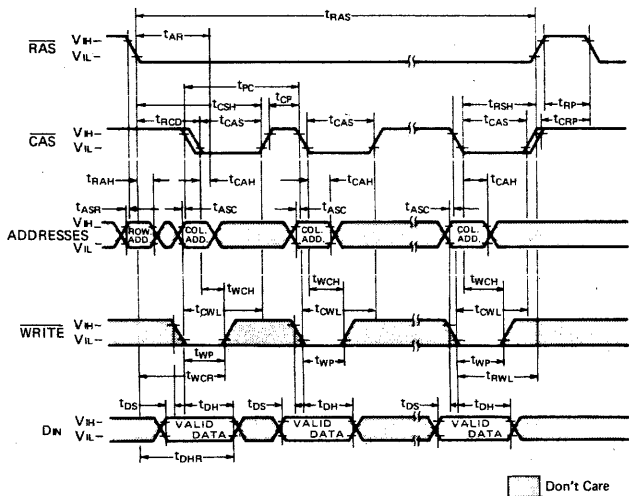
### Read-Write/Read-Modify-Write Cycle



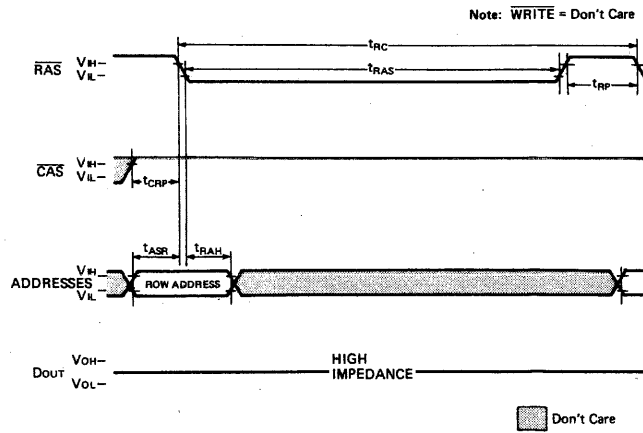
### Page Mode Read Cycle



### Page Mode Write Cycle



### "RAS Only" Refresh Cycle

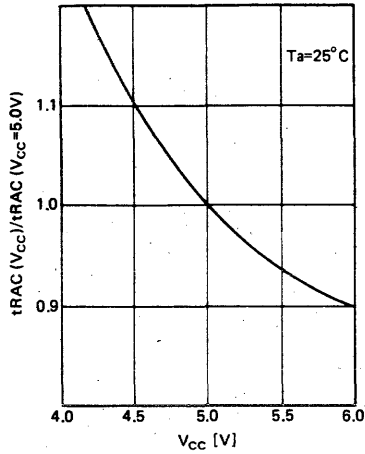




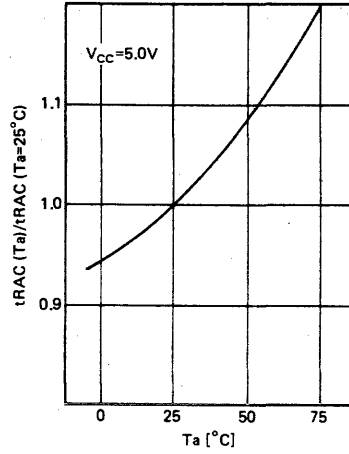
## MN4164, N MOS 65K Bit Dynamic Ram

### Typical Characteristics Curves

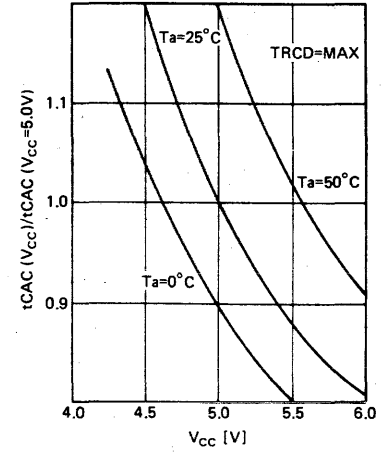
Access time from  $\overline{\text{RAS}}$   
 (Relative value) v.s.  $V_{CC}$



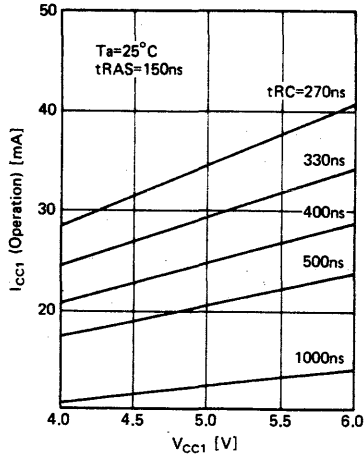
Access time from  $\overline{\text{RAS}}$   
 (Relative value) v.s.  $T_a$



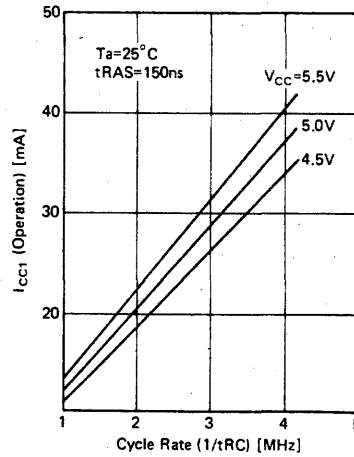
Access time from  $\overline{\text{CAS}}$   
 (Relative value) v.s.  $V_{CC}$



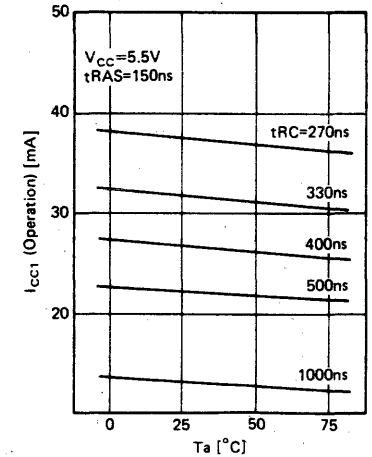
$I_{CC1}$  (tRAS: Constant)  
 v.s.  $V_{CC}$



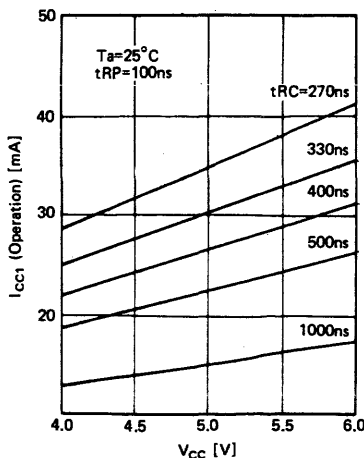
$I_{CC1}$  (tRAS: Constant)  
 v.s. Cycle Rate



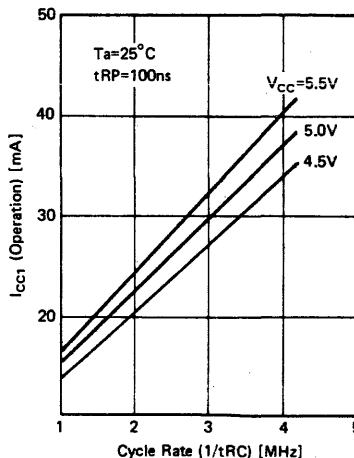
$I_{CC1}$  (tRAS: Constant)  
 v.s.  $T_a$



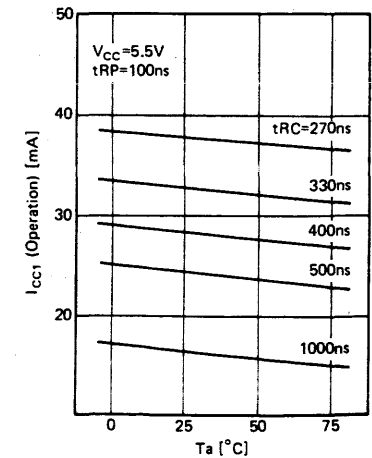
$I_{CC1}$  (tRP: Constant)  
 v.s.  $V_{CC}$



$I_{CC1}$  (tRP: Constant)  
 v.s. Cycle Rate

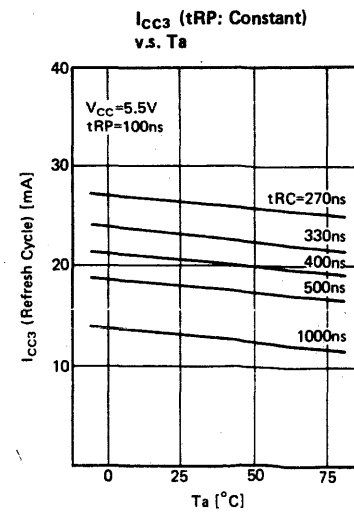
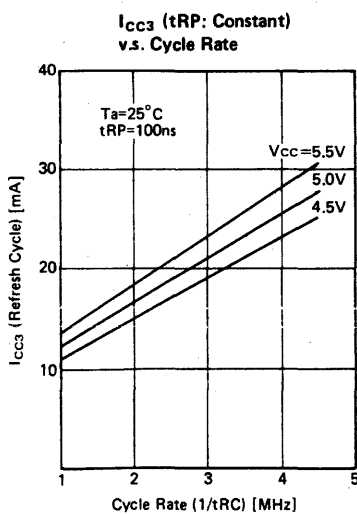
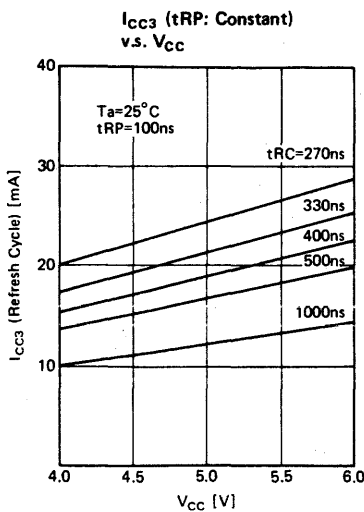
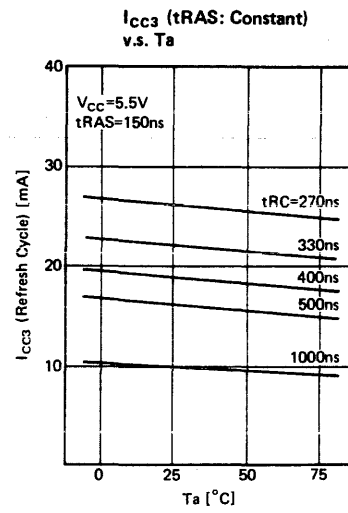
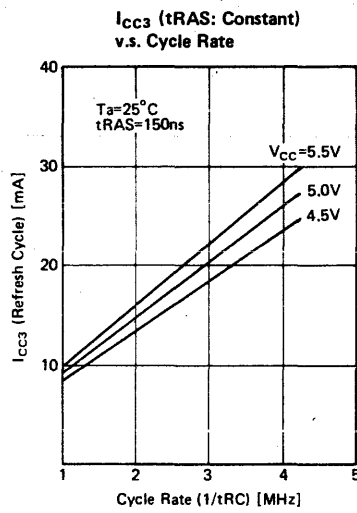
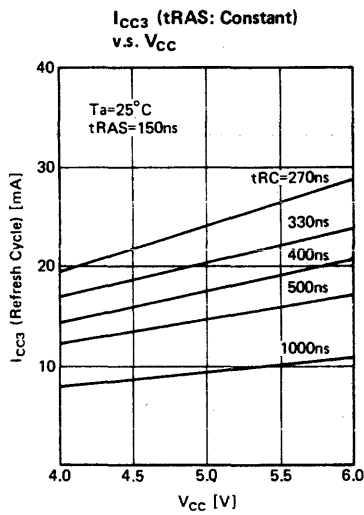
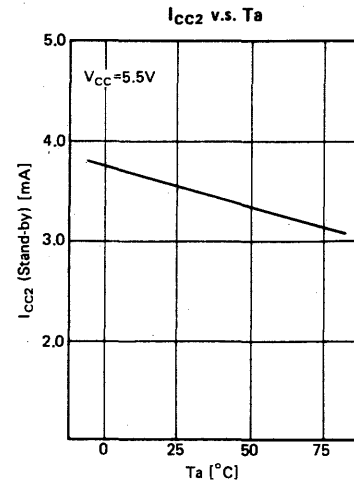
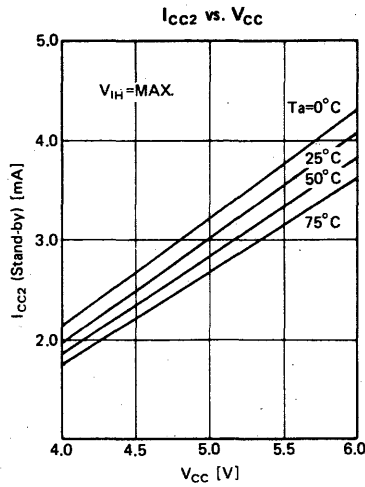


$I_{CC1}$  (tRP: Constant)  
 v.s.  $T_a$



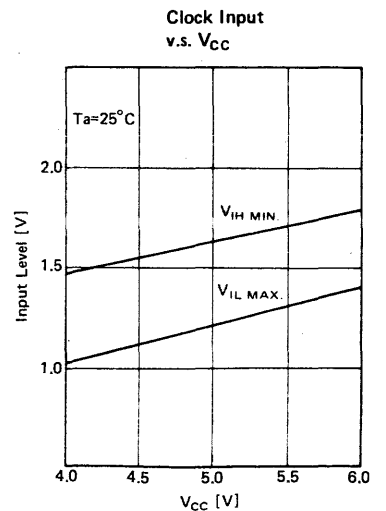
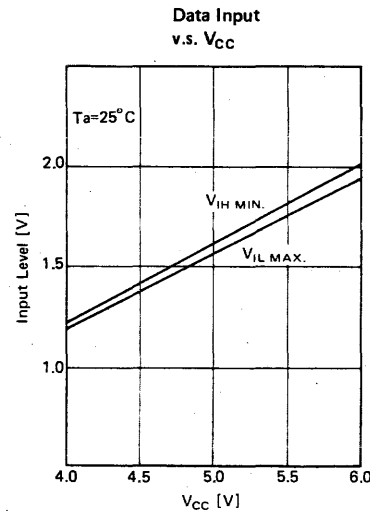
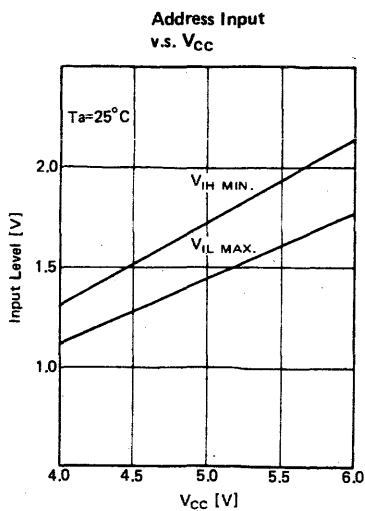
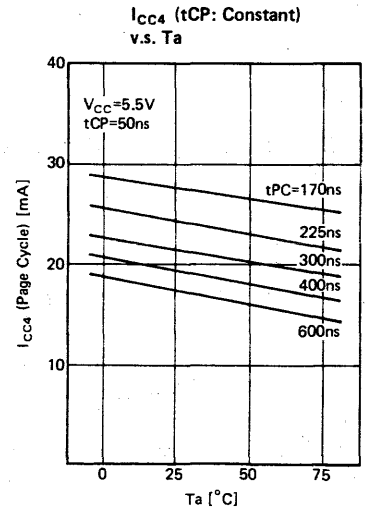
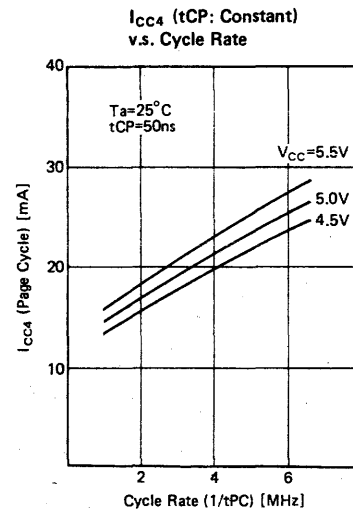
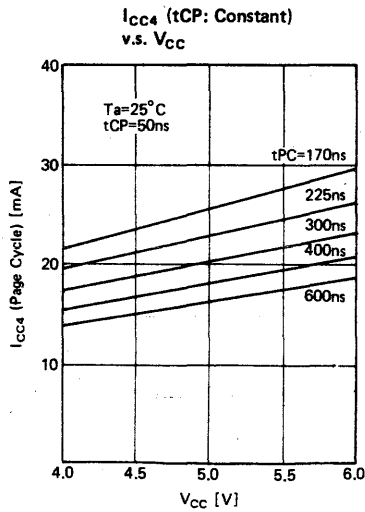
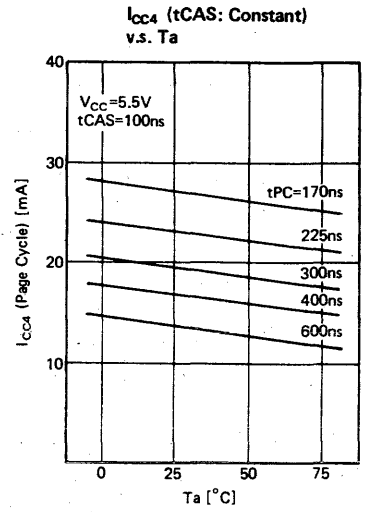
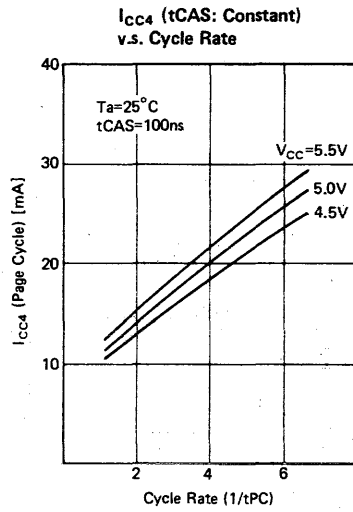
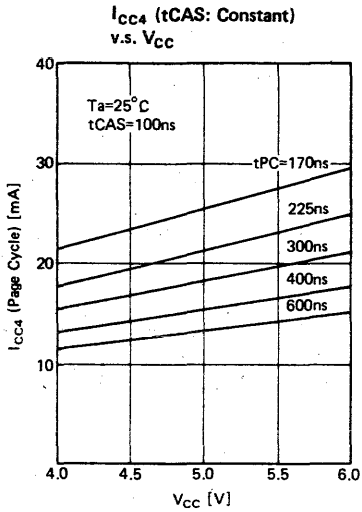
## MN4164, N MOS 65K Bit Dynamic Ram

### Typical Characteristics Curves (Continued)



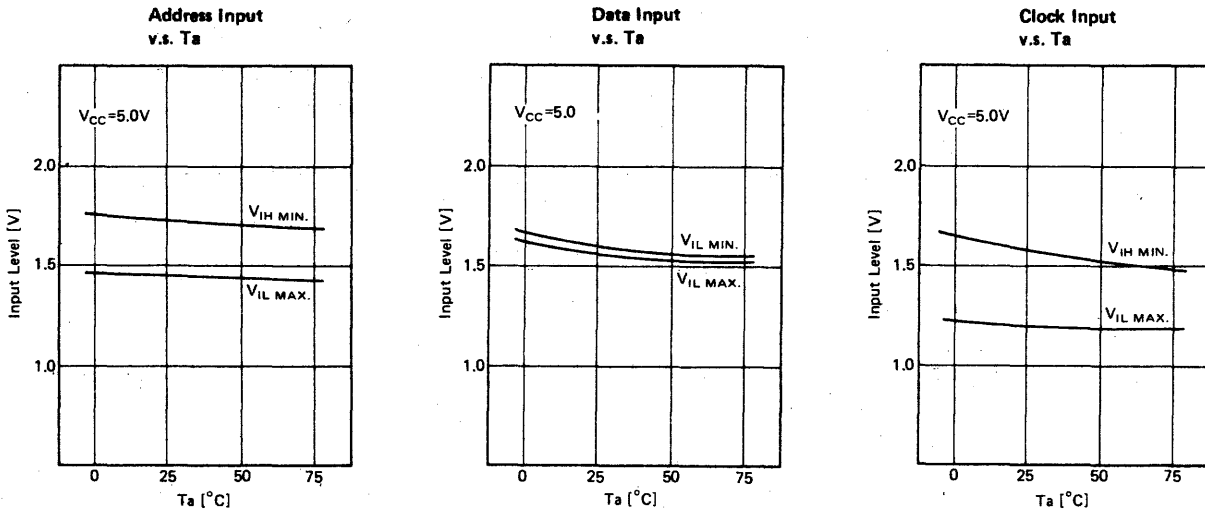
## MN4164, N MOS 65K Bit Dynamic Ram

### Typical Characteristics Curves (Continued)

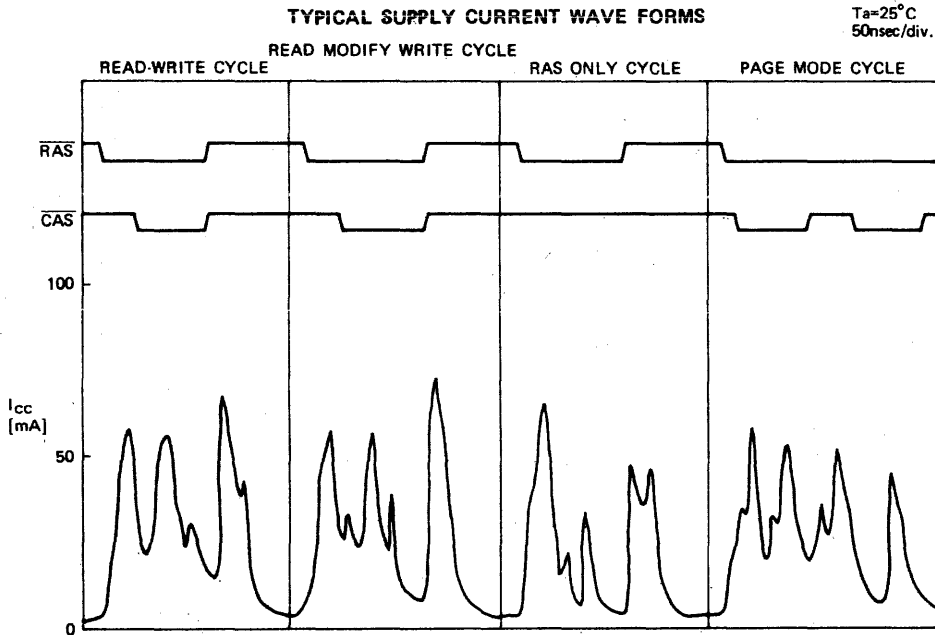


## MN4164, N MOS 65K Bit Dynamic Ram

### Typical Characteristics Curves (Continued)



### Typical Supply Current Wave Forms



Panasonic MEMORY

## MN4164, N MOS 65K Bit Dynamic Ram

### Functional Description

#### Address Input

The 16 address bits required to decode 1 of the 65,536-cell locations within the RAM module are multiplexed onto the 8 address inputs and latched into the on-chip address latches by externally applying two negative going TTL level clocks. The first clock, Row Address Strobe (RAS), latches the 8 row addresses bits into the chip. The second clock, the Column Address Strobe (CAS), subsequently latches the 8 column address bits into the chip. ( $A_0 \sim A_7$  are row and column address input.) Each of these signals, RAS and CAS, triggers a sequence of events which are controlled by different delayed internal clocks. The two clock chains are linked together logically in such a way that the address multiplexing operation is done outside of the critical path timing sequence for read data access. The later events in the CAS clock sequence are inhibited until the occurrence of a delayed signal derived from the RAS clock chain. This "gated CAS" feature allows the CAS clock to be externally activated as soon as the Row Address Hold Time ( $t_{RAH}$ ) has been satisfied and the address inputs have been changed from row address to column address information.

Note that CAS can be activated at any time after  $t_{RAH}$  and it will have no effect on the worst case data access time ( $t_{RAC}$ ) up to the point in time when the delayed row clock no longer inhibits the remaining sequence of column clocks. Two timing specifications result from the internal gating of CAS; they are  $t_{RCD}$  (min) and  $t_{RCD}$  (max). No data storage or reading errors will result if CAS is applied to the RAM module at a point in time beyond the  $t_{RCD}$  (max) limit. However, access time will then be determined exclusively by the access time from CAS ( $t_{CAC}$ ) rather than from RAS ( $t_{RAC}$ ), and access time from RAS will be increased by the amount that  $t_{RCD}$  exceeds the  $t_{RCD}$  (max) limit.

#### Data Input/Output

Data is retrieved from memory in a read cycle by maintaining WRITE in the inactive or high state throughout the portion of the memory cycle in which CAS is active (low). Data read from the selected cell will be available at the output within the specified access time.

Data to be written into a selected cell is latched into an on chip register by a combination of WRITE and CAS while RAS is active. The later of the signals (WRITE or CAS) to make its negative transition is the strobe for the Data in register. This permits several options in the write cycle timing. In a write cycle, if the WRITE input is brought low (active) prior to CAS, the  $D_{IN}$  is strobed by CAS and the set up and hold times are referenced to CAS.

If the input data is not available at CAS time or if it is desired that the cycle be a Read-Modify-Write cycle, the WRITE signal will be delayed until after CAS has made its negative transition. In this delayed write cycle, the data input set up and hold times are referenced to the negative edge of WRITE rather than CAS. (To illustrate this feature,  $D_{IN}$  is referenced to WRITE in the timing diagrams depicting the "Read-Modify-Write" cycle while the "WRITE" and "Page-Mode-Write" cycle diagrams show  $D_{IN}$  referenced to CAS.)

#### Data Output

The normal condition of the data output ( $D_{OUT}$ ) of the RAM module is a high impedance state. That is to say, any time CAS is at a high level, the  $D_{OUT}$  pin will be floating. The only time the outputs will turn on and contain either a logic 0 or logic 1 is at access time during a read cycle.  $D_{OUT}$  will remain valid from access time until CAS is taken back to the inactive (high) conditions.

If the memory cycle in progress is "Read" or "Read-Modify-Write" cycle, then the data output will go from the high impedance state to the active condition and at access time will

contain the data read from the selected cell. This output data is the same polarity (not inverted) as the input data. Once having gone active the output will remain valid until CAS is taken to the precharge state, whether or not RAS goes into precharge.

If the cycle in progress is a "Write" cycle (Write active before CAS goes active), then the output pins will maintain the high impedance throughout the entire cycle.

Note that with this type of output configuration, the user is given full control of the  $D_{OUT}$  pin simply by controlling the placement of WRITE command during write operations, and the pulse width of CAS during read operations. Note also that even though data is not latched at the output, data can remain valid from access time until the beginning of a subsequent cycle without paying any penalty in overall memory cycle time (stretching cycle). This type of output operation results in some very significant system implications.

#### Common I/O

If all write operations are handled in the "Write" cycle mode (WRITE active before CAS goes active), then  $D_{IN}$  can be connected directly to  $D_{OUT}$  respectively for common I/O data bus.

#### RAS and CAS Chip Selection

Only those devices which receive both RAS and CAS signals will execute a read or write cycle. Since  $D_{OUT}$  is not latched, CAS is not required to turn off the outputs of unselected memory device in a matrix. This means that both CAS and/or RAS can be decoded for chip selection. If a common CAS scheme is used where RAS is decoded for module selection, then total memory power can be conserved. If both RAS and CAS are decoded, then a two dimensional (X, Y) chip select array can be realized.

#### Page Mode

The "Page-Mode" feature of the RAM allows for successive memory operations at multiple column locations of the same row address with increased speed without an increase in power. This is done by strobing row address into the chip and maintaining the RAS signal at a logic 0 throughout all successive memory cycle in which the row address is common. This "Page-Mode" of operation will not dissipate the power associated with the negative going edge of RAS. Also, the time required for strobing in a new row address is eliminated, thereby decreasing the access and cycle times.

The successive memory operations in "Page-Mode" may be any sequence of read, write, or read-modify-write operations.

The page boundary of a single RAM module is limited to the 256 column locations determined by all combinations of the 8 column address bits. However, in system applications which utilize more than 65,536 data words (more than one 64K memory block), the page boundary can be extended by using CAS rather than RAS as the chip select signal. RAS is applied to all devices to latch the row address into each device and then CAS is decoded and serves as a page cycle select signal. Only those devices which receive both RAS and CAS signals will execute a read or write cycle.

#### Refresh

Refresh of the dynamic cell matrix is accomplished by performing a memory cycle at each of the 128 row addresses within each 2 ms time interval. Although any normal memory cycle will perform the refresh operation, this function is most easily accomplished with "RAS-Only" cycles. RAS-Only refresh results in a substantial reduction in operating power. This reduction in power is reflected in the  $I_{CC3}$  specification.

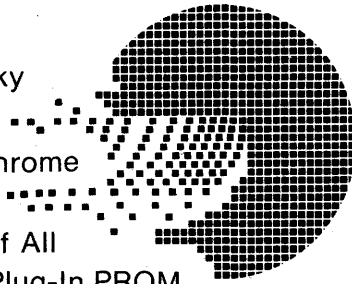


# Field Programmable Read-Only Memories

## 29000 Series

### Features

- Low Power Schottky Technology
- Highly Reliable Nichrome Fuses
- SPROM Versions of All Configurations as Plug-In PROM Replacements
- Typical SPROM "OFF" Power is 25% of Standard Power.
- SPROMs Feature Guaranteed Access Times and Full  $V_{CC}$  Tolerance Under Power-Switched Conditions.
- All Devices Use Same Programming Techniques (Generic)
- All Devices Available in both Commercial ( $0^{\circ}\text{C}$  to  $75^{\circ}\text{C}$ ) and Military ( $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ) Versions.
- Packages Available
  - Flat
  - 0.3 Wide DIP
  - 0.6 Wide DIP
  - 28 Pin LCC

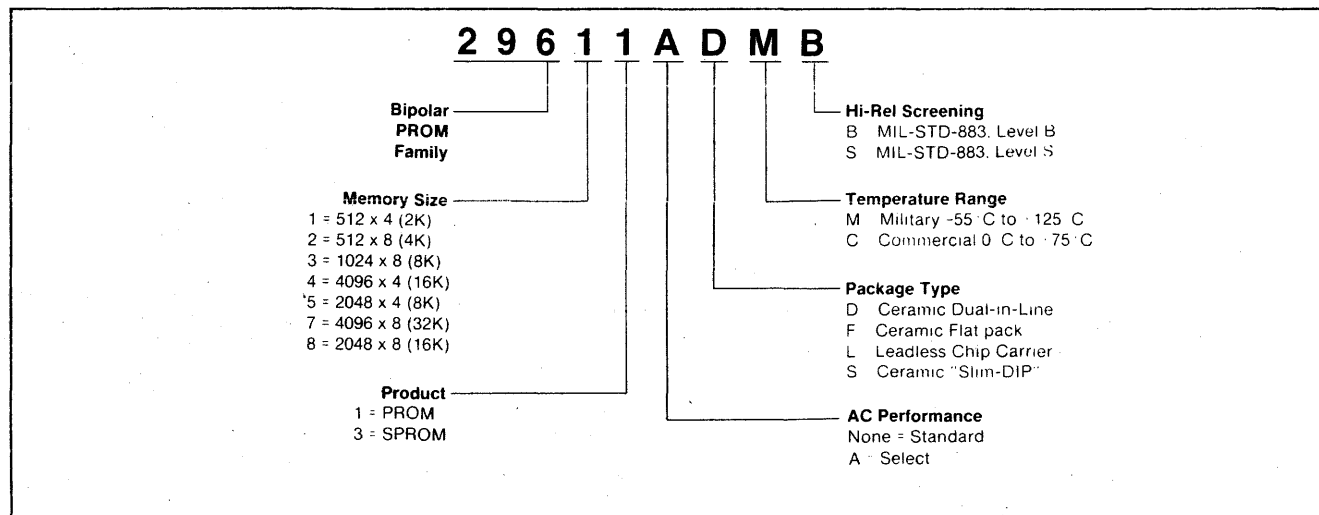


### Absolute Maximum Ratings

(Above which the useful life may be impaired).

- Junction Temperature .....  $-65^{\circ}\text{C}$  to  $+175^{\circ}\text{C}$
- Storage Temperature .....  $-65^{\circ}\text{C}$  to  $+150^{\circ}\text{C}$
- Temperature Under Bias .....  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$
- Supply Voltage to Ground Potential (Continuous) ....  $-0.5\text{V}$  to  $+7.0\text{V}$
- DC Voltage Applied to Outputs (Except During Programming) .....  $-0.5\text{V}$  to  $+V_{CC}$  Max
- DC Input Voltage (Address Inputs) .....  $-0.5\text{V}$  to  $+5.5\text{V}$
- DC Voltage Applied to Outputs During Programming .....  $26\text{V}$
- Output Current into Outputs During Programming .....  $250\text{mA}$
- DC Input Voltage (Chip Select Input-Pin) .....  $-0.5\text{V}$  to  $+33\text{V}$
- DC Input Current .....  $-30\text{mA}$  to  $+5.0\text{mA}$

### Ordering Information



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# 29000 Series PROMs

## Common Electrical Characteristics Over Operating Range (unless otherwise noted)

Parameter	Description	Test Conditions	Min	Typ <sup>1</sup>	Max	Units
V <sub>OH</sub>	Output High Voltage	V <sub>CC</sub> = Min, I <sub>OH</sub> = -2.0mA V <sub>IN</sub> = V <sub>IH</sub> or V <sub>IL</sub>	2.4	3.6		V
V <sub>OL</sub> <sup>3</sup>	Output Low Voltage	V <sub>CC</sub> = Min I <sub>OL</sub> = 8.0mA V <sub>IN</sub> = V <sub>IH</sub> or V <sub>IL</sub> I <sub>OL</sub> = 16mA		0.30 0.35	0.4 0.5	V
V <sub>IH</sub>	Input High Level	Guaranteed Input Logical High Voltage for All Inputs	2.0		5.5	V
V <sub>IL</sub>	Input Low Level	Guaranteed Input Logical Low Voltage for All Inputs	0.0		0.8	V
I <sub>IL</sub>	Input Low Current	V <sub>CC</sub> = Max, V <sub>IN</sub> = 0.4V		-10	-250	μA
I <sub>IH</sub>	Input High Current	V <sub>CC</sub> = Max, V <sub>IN</sub> = 2.7V			10	μA
		V <sub>CC</sub> = Max, V <sub>IN</sub> = 5.5V			0.1	mA
I <sub>SC</sub>	Output Short Circuit Current	V <sub>CC</sub> = Max, V <sub>OUT</sub> = 0.2V (Note 2)	-12	-35	-85	mA
V <sub>I</sub>	Input Clamp Voltage	V <sub>CC</sub> = Min, I <sub>IN</sub> = -18mA			-1.5	V
I <sub>CEx</sub>	Output Leakage Current	V <sub>CC</sub> = Max V <sub>O</sub> = 4.5V Chip Disabled V <sub>O</sub> = 0.45V			+100 -100	μA

Notes: 1. Typical limits are at V<sub>CC</sub> = 5.0V and T<sub>A</sub> = +25°C.

2. Not more than one output should be selected at a time. Duration of the short circuit should not be more than one second.

3. This characteristic cannot be tested prior to programming; it is guaranteed by factory testing.

## SPROM Characteristics

Part Number	Density	Lead Count	I <sub>CC</sub> (mA)		T <sub>AA</sub> (nS) <sup>1</sup>		T <sub>EA</sub> (nS) <sup>1</sup>		T <sub>ER</sub> (nS) <sup>1</sup>	
			Disable	Enable	Mil	Com	Mil	Com	Mil	Com
29633	1024 x 8	24	45	170	90	70	115	75	40	30
29633A	1024 x 8	24	45	170	70	50	70	50	40	30
29653	2048 x 4	18	45	170	90	75	95	80	45	35
29653A	2048 x 4	18	45	170	75	65	80	70	45	35
29673	4096 x 8	24	55	195	105	85	95	125	50	45
29683	2048 x 8	24	50	180	105	85	105	85	50	45
29683A	2048 x 8	24	50	180	70	50	80	65	45	35

## PROM Characteristics

Part Number	Density	Lead Count	I <sub>CC</sub> (mA)	T <sub>AA</sub> (nS) <sup>1</sup>		T <sub>EA</sub> (nS) <sup>1</sup>		T <sub>ER</sub> (nS) <sup>1</sup>	
				Mil	Com	Mil	Com	Mil	Com
29671	4096 x 8	24	195	100	80	50	40	45	40
29671A	4096 x 8	24	195	80		45		35	
29681	2048 x 8	24	180	100	80	50	40	45	40
29681A	2048 x 8	24	180	70	50	45	35	35	30

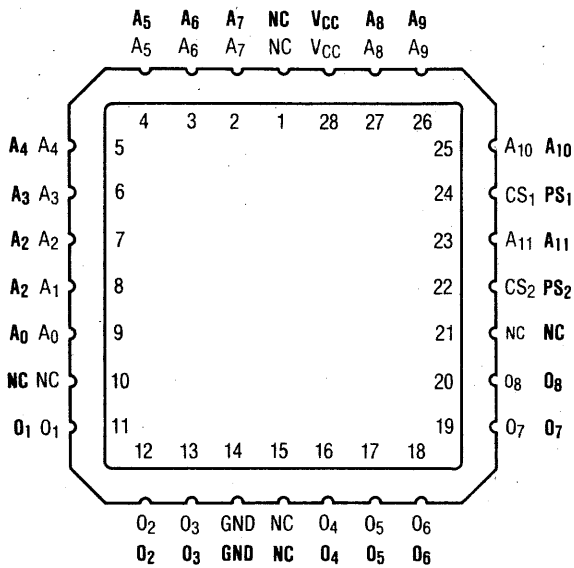
Note: Test Conditions — C<sub>L</sub> = 30pF, R<sub>L</sub> = 300Ω to V<sub>CC</sub> and 600Ω to GND (16mA Load). 300Ω resistor opened for t<sub>EA</sub> and t<sub>ER</sub> measurements between HIGH and OFF states.



## Pin Out Information

29671/A  
29673/A

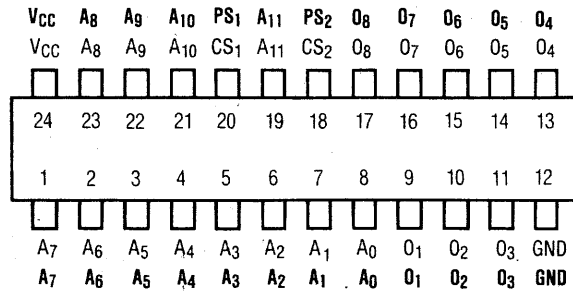
### Leadless Chip Carrier (28 Pin)



Pin 24 is also the programming pin (pp).

29671/A  
29673/A

### Dual-in-Line and Flat Packages

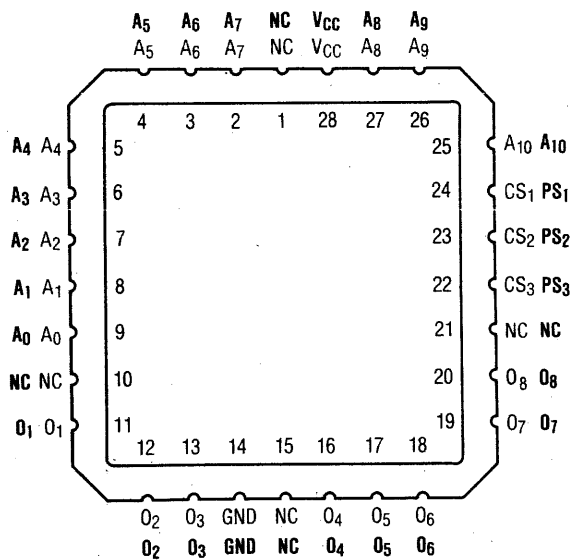


Pin 20 is also the programming pin (pp).

65-01624A

29681/A  
29683/A

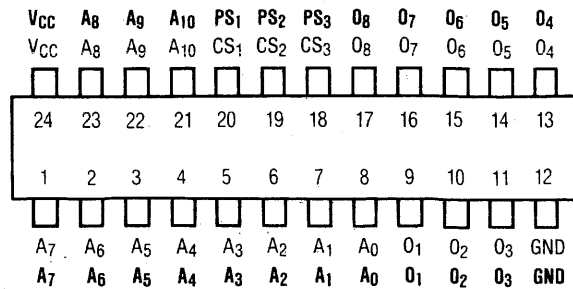
### Leadless Chip Carrier (28 Pin)



Pin 24 is also the programming pin (pp).

29681/A  
29683/A

### Dual-in-Line and Flat Packages



Pin 20 is also the programming pin (pp).

65-01625A

Raytheon MEMORY





### Features

- **Input Latches**
- **Single 5V Supply**
- **Specified over  $V_{CC} \pm 10\%$  Range**
- **TTL Byte Erase/Byte Write/Chip Clear**
- **1 ms or 10 ms Byte Erase/Byte Write**
- **10,000 Erase/Write Cycles per Byte**
- **Silicon Signature™ and DiTrace™**
- **Fast Read Access Time — 250 ns**
- **JEDEC Approved Byte Wide Memory Pinout**

### Description

SEEQ's family of 5 volt, electrically erasable read only memories, E<sup>2</sup>ROMs, include three densities. The 52B13 and 52B13H are 2048 x 8 bit, the 52B23 and 52B23H are 4096 x 8 bit, and the 52B33 and 52B33H are 8192 x 8 bit E<sup>2</sup>ROM. Each device operates on 5 volt TTL levels in the read, write and erase modes. In addition, high voltage (14-22 volts) may be used with the 52B13 for erase and write. All devices have a chip clear mode in which the entire memory is erased in a single erase cycle. The 52B13 performs chip clear with a high voltage signal while

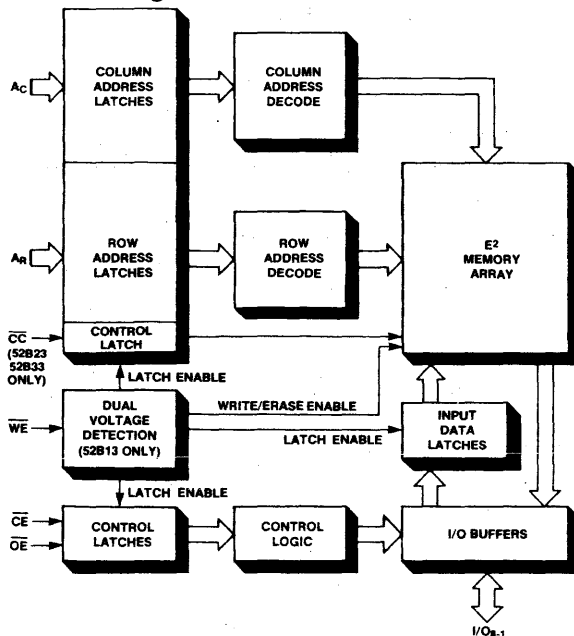
the other devices in the family require only a TTL level signal on  $\overline{OE}$ . The erasure time for both chip clear and byte erase is under 10 ms for all device types. Where faster erase/write times are required, the "H" product may be used to achieve erase or write in under 1 ms.

Data, addresses,  $\overline{CE}$ ,  $\overline{CC}$  and  $\overline{OE}$  are latched on the leading edge of  $\overline{WE}$ . The system controller needs only to maintain the  $\overline{WE}$  signal during the erase/write cycle after the latches are activated. Once written, which requires under 10 ms, there is no limit to the number of times that the data may be read. Each byte may be erased and written at least 10,000 times.

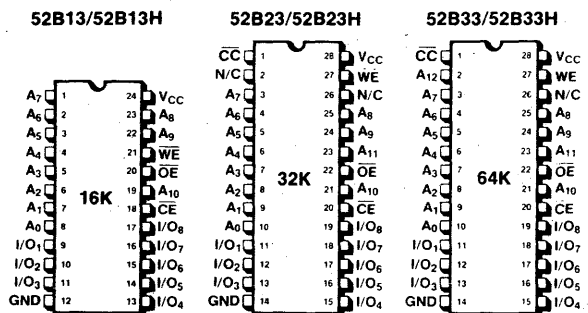
This family of E<sup>2</sup>ROMs include Silicon Signature™ and DiTrace™, SEEQ's on-chip traceability feature. This feature provides the ability to trace the history of any packaged device back to the wafer level. Silicon Signature™ contains device revisions and programming information. The process lot numbers and data is written into DiTrace™.

The members of SEEQ's family of latched E<sup>2</sup>ROMs are each available in the military (-55°C to +125°C) and the extended (-40°C to +85°C) temperature ranges.

### Block Diagram



### Pin Configurations



### Pin Names

A <sub>C</sub>	ADDRESSES — COLUMN (LOWER ORDER BITS)
A <sub>R</sub>	ADDRESSES — ROW
$\overline{CE}$	CHIP ENABLE
$\overline{OE}$	OUTPUT ENABLE
$\overline{WE}$	WRITE ENABLE
I/O	DATA INPUT - WRITE OR ERASE, DATA OUTPUT - READ
$\overline{CC}$	CHIP CLEAR
N/C	NO CONNECT

# 52B13/52B23/52B33 52B13H/52B23H/52B33H

PRELIMINARY DATA SHEET

## Absolute Maximum Stress Ratings\*

### Temperature

Storage ..... -65°C to +100°C

Under Bias ..... -10°C to +80°C

### All Inputs or Outputs with

Respect to Ground ..... +6V to -0.3V

### 52B13 Only

### WE During Writing/Erasing

with Respect to Ground ..... +22.5V to -0.3V

### Duration of WE Supply at

22V During W/E Inhibit ..... 24 Hours

\*COMMENT: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## Recommended Operating Conditions (All Operating Modes — Reference Table 1)

	52BXX-3	52BXX-200/-250/-350
V <sub>CC</sub> Supply Voltage	5V ± 5%	5V ± 10%
Temperature Range	0 to 70°C	0 to 70°C

## D.C. Operating Characteristics During Read or Write/Erase

(Over the Operating V<sub>CC</sub> and Temperature Range)

Symbol	Parameter	Min.	Nom.[1]	Max.	Unit	Test Conditions
I <sub>IN</sub>	Input Leakage Current			10	μA	V <sub>IN</sub> = V <sub>CC</sub> Max.
I <sub>O</sub>	Output Leakage Current			10	μA	V <sub>OUT</sub> = V <sub>CC</sub> Max.
I <sub>WE</sub>	Write Enable Leakage Read Mode			10	μA	WE = V <sub>IH</sub>
	TTL W/E Mode			10	μA	WE = V <sub>IL</sub>
	High Voltage W/E Mode <sup>[2]</sup>			1.5	mA	WE = 22V, CE = V <sub>IL</sub>
	High Voltage W/E Inhibit Mode <sup>[2]</sup>			1.5	mA	WE = 22V, CE = V <sub>IH</sub>
	Chip Erase — TTL Mode			10	μA	WE = V <sub>IL</sub>
	Chip Erase — High Voltage Mode <sup>[2]</sup>			1.5	mA	WE = 22V
I <sub>CC1</sub>	V <sub>CC</sub> Standby Current					
	52B13/H 52B23/H, 52B33/H		15 18	30 40	mA	CE = V <sub>IH</sub>
I <sub>CC2</sub>	V <sub>CC</sub> Active Current					
	52B13/H 52B23/H, 52B33/H		50 60	80 110	mA	CE = OE = V <sub>IL</sub>
V <sub>IL</sub> (D.C.)	Input Low Voltage (D.C.)	-0.1		0.8	V	
V <sub>IL</sub> (A.C.)	Input Low Voltage (A.C.)	-0.4			V	Time = 10 ns
V <sub>IH</sub>	Input High Voltage	2		V <sub>CC</sub> + 1	V	
V <sub>WE</sub>	WE Read Voltage	2		V <sub>CC</sub> + 1	V	
	WE Write/Erase Voltage TTL Mode	-0.1		0.8	V	
	High Voltage Mode <sup>[2]</sup>	14		22	V	
V <sub>OL</sub>	Output Low Voltage			0.45	V	I <sub>OL</sub> = 2.1 mA
V <sub>OH</sub>	Output High Voltage	2.4			V	I <sub>OH</sub> = -400 μA
V <sub>OE</sub>	OE Chip Erase Voltage	14		22	V	I <sub>OE</sub> = 10 μA

### Notes:

1. Nominal values are for T<sub>A</sub> = 25°C and V<sub>CC</sub> = 5.0V.

2. Applies to 52B13 only.

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**52B13/52B23/52B33**  
**52B13H/52B23H/52B33H**  
 PRELIMINARY DATA SHEET

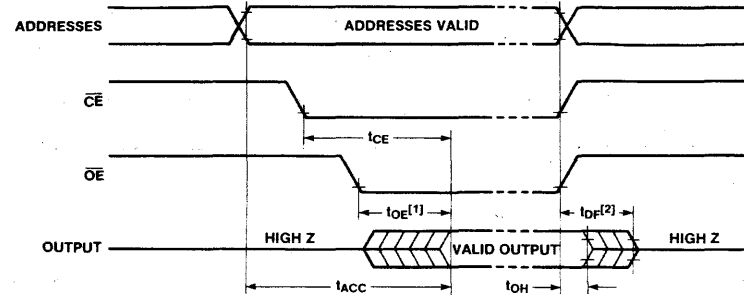
**A.C. Operating Characteristics During Read** (Over Operating V<sub>CC</sub> and Temperature Range)

Symbol	Parameter	Device Number Extension	52BXX/52BXXH			Units	Conditions
			Min.	Nom.	Max.		
t <sub>ACC</sub>	Address to Data Valid	-200			200	ns	$\overline{CE} = \overline{OE} = V_{IL}$
		-250			250	ns	
		-350/-3			350	ns	
t <sub>CE</sub>	Chip Enable to Data Valid	-200			200	ns	$\overline{OE} = V_{IL}$
		-250			250	ns	
		-350/-3			350	ns	
t <sub>OE</sub> <sup>[1]</sup>	Output Enable to Data Valid	-200	10		80	ns	$\overline{CE} = V_{IL}$
		-250	10		90	ns	
		-350/-3	10		100	ns	
t <sub>DF</sub> <sup>[2]</sup>	Output Enable to High Impedance	-200	0		60	ns	$\overline{CE} = V_{IL}$
		-250	0		70	ns	
		-350/-3	0		80	ns	
t <sub>OH</sub>	Output Hold				0	ns	$\overline{CE} = \overline{OE} = V_{IL}$
C <sub>IN</sub> / C <sub>OUT</sub> <sup>[3]</sup>	Input/Output Capacitance				10	pF	V <sub>IN</sub> =0/V <sub>OUT</sub> =0, T <sub>A</sub> = 25°C

**A.C. Operating Characteristics During Write/Erase** (Over Operating V<sub>CC</sub> and Temperature Range)

Symbol	Parameter	Min.	Max.	Units
Q <sup>[4]</sup>	Maximum Endurance	10,000		Cycles/Byte
t <sub>s</sub>	$\overline{CC}$ , $\overline{CE}$ , $\overline{OE}$ or A <sub>N</sub> Setup to $\overline{WE}$	50		ns
t <sub>DS</sub>	Data Setup to $\overline{WE}$	0		ns
t <sub>H</sub> <sup>[5]</sup>	$\overline{WE}$ to $\overline{CE}$ , $\overline{OE}$ , $\overline{CC}$ , A <sub>N</sub> or Data Change	50		ns
t <sub>WP</sub>	Write Enable, $\overline{WE}$ , Pulse Width	Chip Clear — All Devices	9	ms
		Byte Modes — 52BXX	9	ms
		Byte Modes — 52BXXH	1	ms
t <sub>WR</sub> <sup>[6]</sup>	$\overline{WE}$ to Mode Change	50		ns

**READ TIMING**

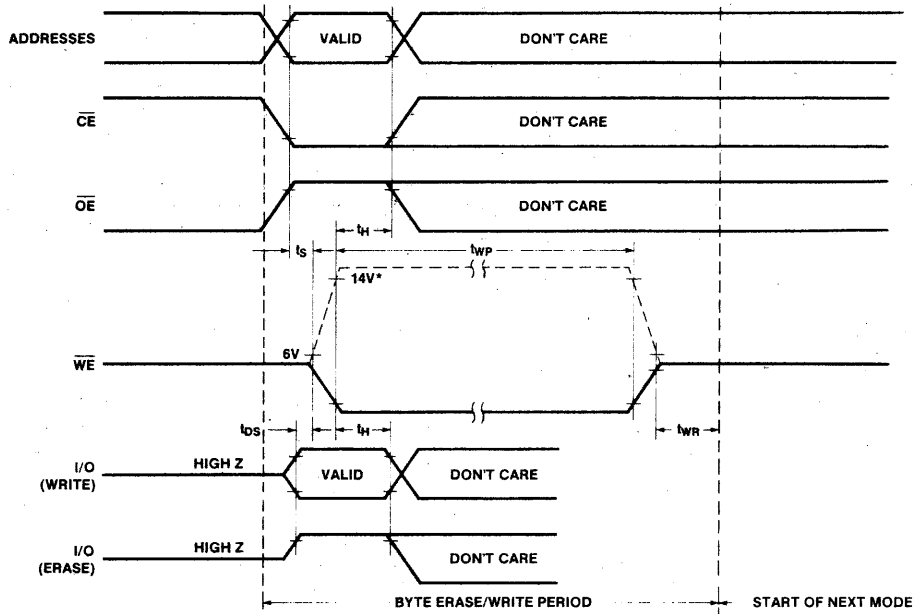


- Notes:
1.  $\overline{OE}$  may be delayed up to t<sub>ACC</sub> — t<sub>OE</sub> after the falling edge of  $\overline{CE}$  without impact on t<sub>ACC</sub>.
  2. t<sub>DF</sub> is specified from  $\overline{OE}$  or  $\overline{CE}$ , whichever occurs first.
  3. This parameter is periodically sampled.
  4. Maximum endurance, Q, is the number of write and erase cycles/byte.
  5. After t<sub>H</sub>, hold time, from  $\overline{WE}$ , the inputs  $\overline{CE}$ ,  $\overline{OE}$ ,  $\overline{CC}$ , Address and Data are latched and are "Don't Cares" until t<sub>WR</sub>, Write Recovery Time, after the trailing edge of  $\overline{WE}$ .
  6. The Write Recovery Time, t<sub>WR</sub>, is the time after the trailing edge of  $\overline{WE}$  that the latches are open and able to accept the next mode set-up conditions. Reference Table 1 (page 2) for mode control conditions.

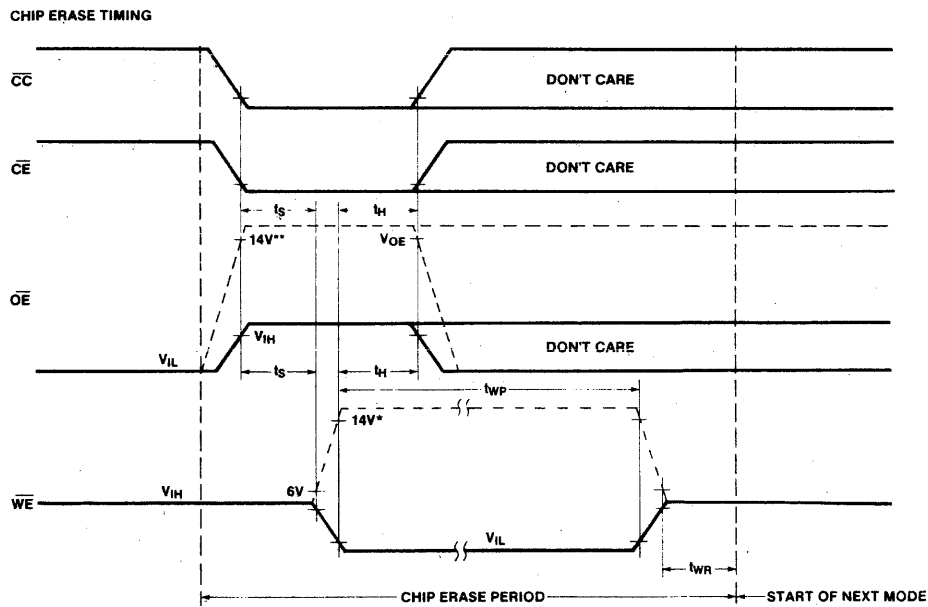
**52B13/52B23/52B33**  
**52B13H/52B23H/52B33H**

PRELIMINARY DATA SHEET

**BYTE ERASE OR BYTE WRITE TIMING**



**CHIP ERASE TIMING**



ADDRESS AND I/O PINS ARE "DON'T CARE".

\*IN THE BYTE ERASE/WRITE OR CHIP ERASE MODE, WE FOR THE 52B13 MAY BE EITHER A HIGH OR TTL VOLTAGE. WE FOR THE 52XXH IS TTL ONLY.

\*\*IN THE CHIP ERASE MODE, OE IS A HIGH VOLTAGE FOR THE 52B13/52B13H ONLY.

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### Features

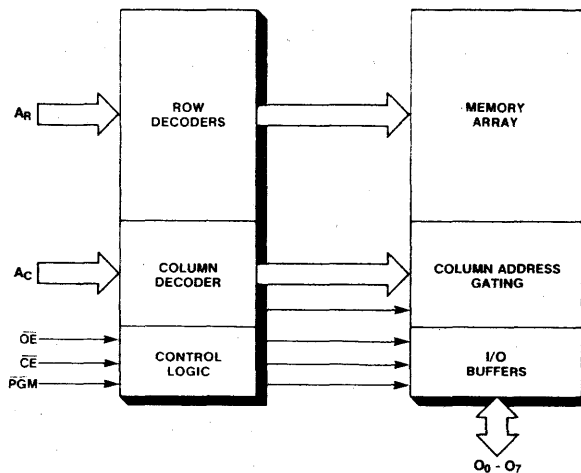
- **200 ns Access Times at 0 to 70° C**
- **Programmed Using Intelligent Algorithm**
  - Typically 5 ms/byte Programming Time
  - 2 Minutes for 5143 (27128)
  - 1 Minute for 5133 (2764)
- **JEDEC Approved Byte-wide Pin Configuration**
  - 5133 8K x 8 Organization
  - 5143 16K x 8 Organization
- **Low Power Dissipation**
  - 100 mA Active Current
  - 30 mA Standby Current
- **Extended Temperature Range Available**
- **Silicon Signature™**

### Description

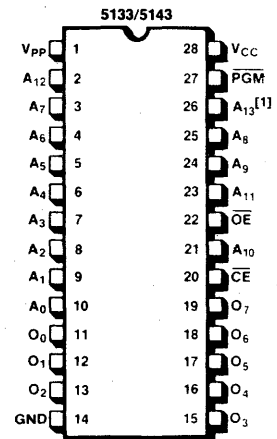
SEEQ's 5133 and 5143 are ultraviolet light erasable EPROMs which are organized 8K x 8 and 16K x 8 respectively. They are pin for pin compatible to 64K and 128K EPROMs, such as the 2764 and 27128 respectively, in all operational/programming modes. Both devices have access times as fast as 200 ns over the 0 to 70° C temperature and  $V_{CC}$  tolerance range. The access time is achieved without sacrificing power since the maximum active and standby currents are 100 mA and 30 mA respectively. The 200 ns allows higher system efficiency by eliminating the need for wait states in today's 8- or 16-bit microcomputers.

Initially, and after erasure, all bits are in the "1" state. Data is programmed by applying 21 V to  $V_{PP}$  and a TTL "0" to pin 27 (program pin). The 5133 and 5143 may be programmed with an intelligent algorithm (continued on page 2)

### Block Diagram



### Pin Configuration



NOTE 1: PIN 26 IS A NO CONNECT ON THE 5133.

### Mode Selection

MODE \ PINS	CE (20)	OE (22)	PGM (27)	V <sub>PP</sub> (1)	V <sub>CC</sub> (28)	Outputs (11-13, 15-19)
Read	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>CC</sub>	V <sub>CC</sub>	D <sub>OUT</sub>
Standby	V <sub>IH</sub>	X	X	V <sub>CC</sub>	V <sub>CC</sub>	High Z
Program	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>IL</sub>	V <sub>PP</sub>	V <sub>CC</sub>	D <sub>IN</sub>
Program Verify	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>PP</sub>	V <sub>CC</sub>	D <sub>OUT</sub>
Program Inhibit	V <sub>IH</sub>	X	X	V <sub>PP</sub>	V <sub>CC</sub>	High Z
Silicon Signature™	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>CC</sub>	V <sub>CC</sub>	Encoded Data

X can be either V<sub>IL</sub> or V<sub>IH</sub>

For Silicon Signature™: A<sub>0</sub>-A<sub>3</sub> are toggled. A<sub>4</sub> = V<sub>IL</sub>. A<sub>9</sub> = 12V. all other addresses are at any TTL level.

### Pin Names

A <sub>C</sub>	ADDRESSES — COLUMN (LSB)
A <sub>R</sub>	ADDRESSES — ROW
CE	CHIP ENABLE
OE	OUTPUT ENABLE
O <sub>0</sub> - O <sub>7</sub>	OUTPUTS
PGM	PROGRAM

that is now available on commercial programmers. The programming time is typically 5 ms/byte or 2 minutes for all 16K bytes of the 5143. The 5133 requires only half of this time, about a minute for 8K bytes. This faster time improves manufacturing throughput time by hours over conventional 50 ms algorithms. Commercial programmers (e.g. Data I/O, Pro-log, Digelec, Kontron, and Stag) have implemented this fast algorithm for SEEQ's EPROMs. If desired, the 5143 and the 5133 may be programmed

using the conventional 50 ms programming specification of older generation EPROMs.

Incorporated on the 5143 and 5133 is Silicon Signature™. Silicon Signature contains encoded data which identifies SEEQ as the EPROM manufacturer, the product's fab location, and programming information. This data is encoded in ROM to prevent erasure by ultraviolet light.

**Absolute Maximum Stress Ratings**

Temperature

Storage ..... -65° C to +150° C

Under Bias ..... -10° C to +80° C

All Inputs or Outputs with

Respect to Ground ..... +7V to -0.6V

V<sub>PP</sub> During Programming with

Respect to Ground ..... +22V to -0.6V

Voltage on A<sub>0</sub> with

Respect to Ground ..... +15.5V to -0.6V

\*COMMENT: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Recommended Operating Conditions** (51XX = 5133 and 5143)

	51XX-200, 51XX-250 51XX-300, 51XX-450	51XX-2, 51XX-3, 51XX-4
V <sub>CC</sub> Supply Voltage <sup>1</sup>	5 V ± 10%	5 V ± 5%
Temperature Range (Read Mode)	0 to 70° C	0 to 70° C
V <sub>PP</sub> During Programming	21 ± 0.5 V	21 ± 0.5 V

**DC Operating Characteristics During Read or Programming**

Symbol	Parameter	Limits		Unit	Test Conditions
		Min.	Max.		
I <sub>IN</sub>	Input Leakage Current		10	μA	V <sub>IN</sub> = V <sub>CC</sub> Max.
I <sub>O</sub>	Output Leakage Current		10	μA	V <sub>OUT</sub> = V <sub>CC</sub> Max.
I <sub>PP</sub> <sup>2</sup>	V <sub>PP</sub> Current	Read Mode	5	mA	V <sub>PP</sub> = V <sub>CC</sub> Max.
		Prog. Mode	30	mA	V <sub>PP</sub> = 21.5V
I <sub>CC1</sub> <sup>2</sup>	V <sub>CC</sub> Standby Current		30	mA	$\overline{CE} = V_{IH}$
I <sub>CC2</sub> <sup>2</sup>	V <sub>CC</sub> Active Current		100	mA	$\overline{CE} = \overline{OE} = V_{IL}$
V <sub>IL</sub>	Input Low Voltage	-0.1	0.8	V	
V <sub>IH</sub>	Input High Voltage	2	V <sub>CC</sub> + 1	V	
V <sub>OL</sub>	Output Low Voltage		0.45	V	I <sub>OL</sub> = 2.1 mA
V <sub>OH</sub>	Output High Voltage	2.4		V	I <sub>OH</sub> = -400 μA

**NOTES:**

1. V<sub>CC</sub> must be applied simultaneously or before V<sub>PP</sub> and removed simultaneously or after V<sub>PP</sub>.

**AC Operating Characteristics During Read**

Symbol	Parameter	Limits (nsec)								Test Conditions
		51XX-2 51XX-200		51XX-250		51XX-3 51XX-300		51XX-4 51XX-450		
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
$T_{ACC}$	Address to Data Valid		200		250		300		450	$\overline{CE} = \overline{OE} = V_{IL}$
$T_{CE}$	Chip Enable to Data Valid		200		250		300		450	$\overline{OE} = V_{IL}$
$T_{OE}$	Output Enable to Data Valid		75		100		120		150	$\overline{CE} = V_{IL}$
$T_{DF}$	Output Enable to Output Float	0	60	0	60	0	105	0	130	$\overline{CE} = V_{IL}$
$T_{OH}$	Output Hold from Chip Enable, Addresses, or Output Enable whichever occurred first	0		0		0		0		$\overline{CE} = \overline{OE} = V_{IL}$

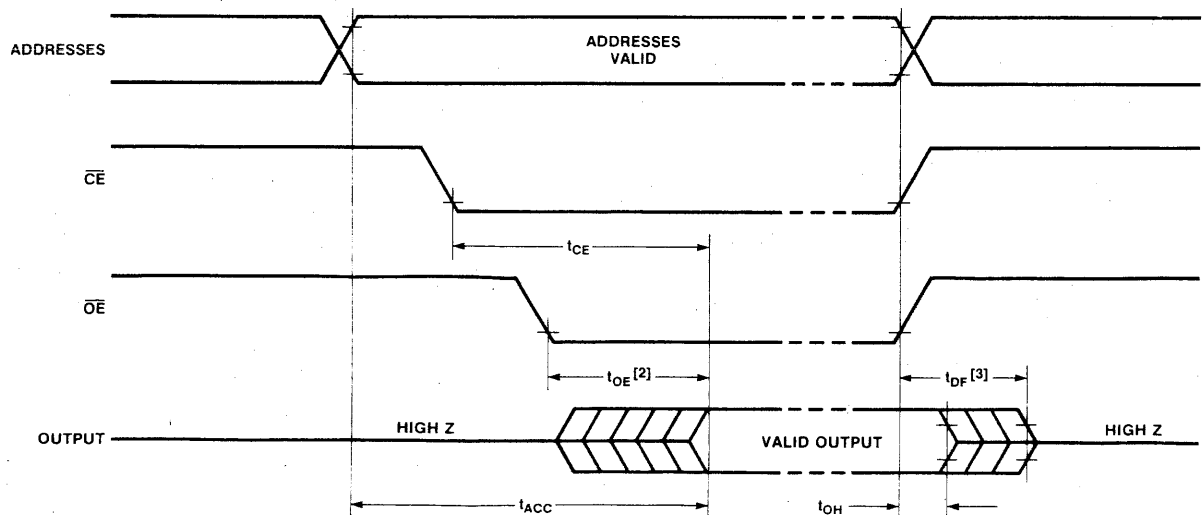
**Capacitance<sup>[1]</sup>**

Symbol	Parameter	Typ.	Max.	Unit	Conditions
$C_{IN}$	Input Capacitance	4	6	pF	$V_{IN} = 0V$
$C_{OUT}$	Output Capacitance	8	12	pF	$V_{OUT} = 0V$

**A.C. Test Conditions**

Output Load: 1 TTL gate and  $C_L = 100$  pF  
 Input Rise and Fall Times:  $\leq 20ns$   
 Input Pulse Levels: 0.45V to 2.4V  
 Timing Measurement Reference Level:  
 Inputs 1V and 2V  
 Outputs 0.8V and 2V

**A.C. Waveforms**



- NOTES:  
 1. THIS PARAMETER IS SAMPLED AND IS NOT 100% TESTED.  
 2.  $\overline{OE}$  MAY BE DELAYED UP TO  $t_{ACC} - t_{OE}$  AFTER THE FALLING EDGE OF  $\overline{CE}$  WITHOUT IMPACT ON  $t_{ACC}$ .  
 3.  $t_{DF}$  IS SPECIFIED FROM  $\overline{OE}$  OR  $\overline{CE}$ , WHICHEVER OCCURS FIRST.

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## BIPOLAR MEMORY Products

### Integrated Fuse Logic (IFL)

20 Pin: 82S151  
82S153/82S153A(T.S.)  
82S159(T.S.)

24 Pin: 82S167(T.S.)

28 Pin: 82S100(T.S.)/82S101(O.C.)  
82S102(O.C.)/82S103(T.S.)  
82S105A(T.S.)

### RAMs

256-BIT: 82LS16(T.S.)/82LS17(O.C.)

2034-BIT: 82S212A(T.S.)

### PROMs

256-BIT: 82S23A(O.C.)/82S123A(T.S.)

1024-BIT: 82S126A(O.C.)/82S129A(T.S.)

2048-BIT: 82S130A(O.C.)/82S131A(T.S.)  
82LS135(T.S.)

4096-BIT: 82S137A/82S137B(T.S.)  
82S147A(T.S.)

8192-BIT: 82S181A/82S181B(T.S.)  
82S185A/82S185B(T.S.)

16,384-BIT: 82S191A(T.S.)  
82HS195(T.S.)

32,768-BIT: 82HS321(T.S.)

# signetics

a subsidiary of U.S. Phillips Corporation

Signetics Corporation  
811 E. Arques Avenue  
P.O. Box 3409  
Sunnyvale, California 94088-3409  
Telephone 408/739-7700

Signetics

MEMORY

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**SELECTION GUIDE**


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Device	Org	Output CKT 1	Output Logic 2	Access Time 3	Temp 4	Pkg 5	Pins	Max Icc
<b>FPLAs</b>								
82S100	16x48x8	TS	—	50	M,C	FN	28	170
82S101	16x48x8	OC	—	50	M,C	FN	28	170
82S152	18x32x10	OC	I/O	40	M,C	FN	20	155
82S152A*	18x32x10	OC	I/O	30	M,C	FN	20	155
82S153	18x32x10	TS	I/O	40	M,C	FN	20	155
82S153A*	18x32x10	TS	I/O	30	M,C	FN	20	155
82S161*	12x48x8	TS	—	50	M,C	FN	24	170
82S173*	20x24x11	TS	I/O	25	M,C	FN	24	170
<b>FPGAs</b>								
82S151*	18x12	TS	I/O	20	M,C	FN	20	155
82S103	16x9	TS	—	35	M,C	FN	28	170
<b>FPLSs</b>								
82S105	16x48x8	TS	R	60	M,C	FN	28	180
82S105A	16x48x8	TS	R	50	M,C	FN	28	180
82S155*	16x32x12	TS	I/O,R	65	M,C	FN	20	155
82S157*	16x32x12	TS	I/O,R	65	M,C	FN	20	155
82S159	16x32x12	TS	I/O,R	65	M,C	FN	20	155
82S167	14x48x6	TS	R	65	C	F3,N3	24	160
<b>PROMs</b>								
82S23	32x8	OC	—	50	M,C	FN	16	77
82S23A*	32x8	OC	—	25	M,C	FN	16	100
82S123	32x8	TS	—	50	M,C	FN	16	77
82S123A*	32x8	TS	—	25	M,C	FN	16	100
82S126	256x4	OC	—	50	M,C	FN	16	120
82S126A*	256x4	OC	—	35	M,C	FN	16	120
82S129	256x4	TS	—	50	M,C	FN	16	120
82S129A*	256x4	TS	—	35	M,C	FN	16	120
10149	256x4	OE	—	20	C	F	16	150
100149	256x4	OE	—	20	C	F	16	150
82S130	512x4	OC	—	50	M,C	FN	16	140
82S130A*	512x4	OC	—	35	M,C	FN	16	140
82S131	512x4	TS	—	50	M,C	FN	16	140
82S131A*	512x4	TS	—	35	M,C	FN	16	140
82LS135	256x8	TS	—	100	M,C	FN	20	100
82S135*	256x8	TS	—	45	M,C	FN	20	155
82S115	512x8	TS	L	60	M,C	FN	24	175
82S141	512x8	TS	—	60	M,C	FN	24	175
82S137	1024x4	TS	—	60	M,C	FN	18	140
82S137A	1024x4	TS	—	45	M,C	FN	18	140
82S137B	1024x4	TS	—	35	C	FN	18	140
82S147	512x8	TS	—	60	M,C	FN	20	155
82S147A	512x8	TS	—	45	M,C	FN	20	155
82LS181	1024x8	TS	—	150	M,C	FN	24	80
82S180	1024x8	OC	—	70	M,C	FN	24	175
82S181	1024x8	TS	—	70	M,C	FN	24	175
82S181A	1024x8	TS	—	50	M,C	FN	24	175
82S181B	1024x8	TS	—	45	C	FN	24	175
82S183	1024x8	TS	L	60	M,C	FN	24	175
82S2708	1024x8	TS	—	90	M	F,R,G	24	185
82S185	2048x4	TS	—	100	M,C	FN	18	120
82S185A	2048x4	TS	—	50	M,C	FN	18	155
82S185B	2048x4	TS	—	45	C	FN	18	155
82S191	2048x8	TS	—	80	M,C	FN	24	175
82S191A	2048x8	TS	—	55	M,C	FN	24	175
82S191B*	2048x8	TS	—	45	M,C	FN	24	175
82HS195*	4096x4	TS	—	30	M,C	FN	20	155
82S321*	4096x8	TS	—	70	M,C	FN	24	175
82HS321*	4096x8	TS	—	40	M,C	FN	24	175
82HS641*	8192x8	TS	—	45	M,C	FN	24	175

## SELECTION GUIDE

Device	Org	Output CKT 1	Output Logic 2	Access Time 3	Temp 4	Pkg 5	Pins	Max Icc
<b>CAMs</b>								
10155	8x2	OE	—	13	C	F,N	18	140
<b>RAMs</b>								
3101A	16x4	OC	B	35	C	F,N	16	105
54/74S189	16x4	TS	B	35	M,C	F,N	16	110
82S21	32x2	OC	T	50	C	F,N	16	130
82S16	256x1	TS	T	50	M,C	F,N	16	115
82S17	256x1	OC	T	50	M,C	F,N	16	115
74S301	256x1	OC	B	50	M,C	F,N	16	115
82LS16	256x1	TS	T	40	M,C	F,N	16	70
82LS17	256x1	OC	T	40	M,C	F,N	16	70
74LS301	256x1	OC	B	40	M,C	F,N	16	70
82S09	64x9	OC	T	45	M,C	F,N	28	190
82S09A	64x9	OC	T	35	C	F,N	28	190
82S19	64x9	OC	B	35	M,C	F,N	28	190
82S210	256x9	TS	B	60	M,C	F,N	24	185
82S212	256x9	TS	B	45	M,C	F,N	22	185
82S212A	256x9	TS	B	35	C	F,N	22	185
8X350	256x8	TS	B	N/A	M,C	F,N	22	185

**NOTES:**

- 1 Output circuit  
OE = Open Emitter  
OC = Open Collector  
TS = 3 State
- 2 Output logic  
T = Transparent input data appears on output during Write  
B = Blanked output is blanked during Write  
R = Register  
I/O = Input/Output option  
L = Latch
- 3 Commercial (0°C to +75°C) except (82S2708)
- 4 Temperature range  
C = Commercial (0°C to +75°C)  
M = Military (-55°C to +125°C)  
A = Automotive (-40°C to +85°C)
- 5 Packages:  
F = Hermetic Cerdip Dual In Line  
N = Plastic Dual In Line  
R = Ceramic Flat Pack (Military only)  
G = Ceramic Square Leadless Chip Carrier (Military only)

\*Under development

**FIELD PROGRAMMABLE GATE ARRAY (18x15x12)**

**82S151**

INTEGRATED FUSE LOGIC  
SERIES 20

**DESCRIPTION**

The 82S151 is a single level logic element, consisting of 12 AND gates with fusible link connections for programming I/O polarity, I/O direction and output enable control.

All gates are linked to 6 inputs (I) and 12 bidirectional I/O lines (B). These yield variable I/O gate configurations via 3 direction control gates (D), ranging from 18 inputs to 12 outputs.

On chip T/C buffers couple either True (I, B) or Complement ( $\bar{I}$ ,  $\bar{B}$ ) input polarities to each AND gate. The polarity of all gate outputs is individually programmable through a set of EX-OR gates for implementing AND/NAND logic functions. Alternately, if desired, OR/NOR logic functions can also be realized by programming for each gate the complement of its inputs and output (DeMorgan's Theorem).

The 82S151 is field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

This device is available in a 20-pin slim-line package. For the commercial temperature range (0°C to +75°C) specify N82S151 N or F. For the military temperature range (-55°C to +125°C) specify S82S151 F only.

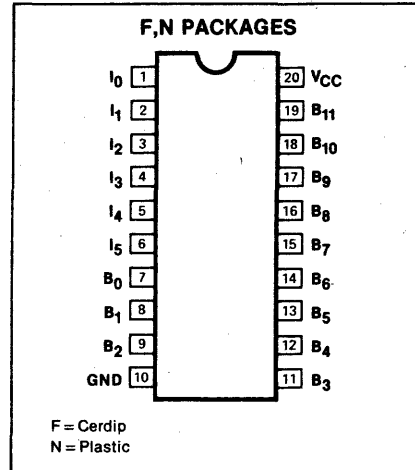
**FEATURES**

- Field Programmable (Ni-Cr link)
- 6 inputs
- 15 Product Terms:
  - 12 Logic Terms
  - 3 Control Terms
- 12 bidirectional I/O lines
- Active high or low outputs
- Programmable output enable
- Power dissipation: 650mW (typ)
- I/O propagation delay: 25ns (max)
- Input loading
  - N82S151: -100µA (max)
  - S82S151: -150µA (max)
- Output: three-state
- TTL compatible

**APPLICATIONS**

- Random gating functions
- Address decoding
- Code detectors
- Memory mapped I/O
- Fault monitors
- I/O port decoders

**PIN CONFIGURATION**



**LOGIC FUNCTIONS**

Typical Output Functions:

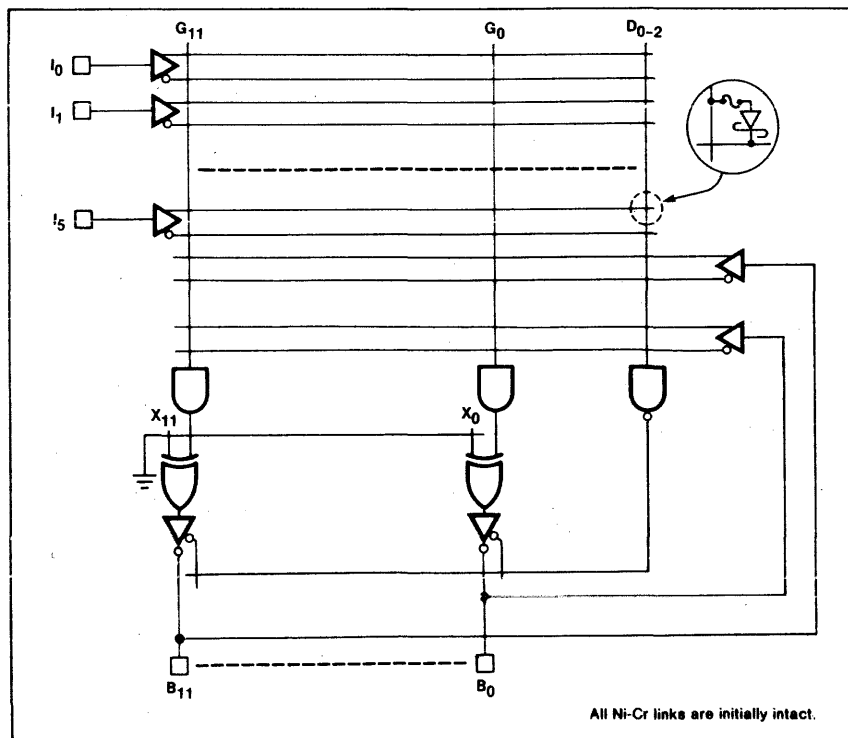
Active-High  
 $X = A \cdot \bar{B} \cdot C \cdot \dots$

Active-Low  
 $X = A \cdot \bar{B} \cdot C \cdot \dots$   
 $X = \bar{A} + B + \bar{C} + \dots$

NOTES:

- For each of the 12 outputs, either function X (active-high) or  $\bar{X}$  (active-low) is available, but not both. The desired output polarity is programmed via the EX-OR gates.
- X, A, B, C, etc. are user defined connections to fixed inputs (I) and bidirectional pins (B).

**FUNCTIONAL DIAGRAM**



MEMORY Signetics

# FIELD PROGRAMMABLE GATE ARRAY (18x15x12)

## 82S151

INTEGRATED FUSE LOGIC  
SERIES 20

### ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING <sup>1</sup>		UNIT
	Min	Max	
V <sub>CC</sub>	Supply voltage	+7	Vdc
V <sub>IN</sub>	Input voltage	+5.5	Vdc
V <sub>OUT</sub>	Output voltage	+5.5	Vdc
I <sub>IN</sub>	Input currents	-30	mA
I <sub>OUT</sub>	Output currents	+100	mA
T <sub>A</sub>	Temperature range		C°
T <sub>STG</sub>	Operating		
	N82S151	0	+75
T <sub>STG</sub>	S82S151	-55	+125
	Storage	-65	+150

### THERMAL RATINGS

TEMPERATURE	Milli-tary	Commer-cial
Maximum junction	175°C	150°C
Maximum ambient	125°C	75°C
Allowable thermal rise ambient to junction	50°C	75°C

### DC ELECTRICAL CHARACTERISTICS

N82S151: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S151: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TEST CONDITION	N82S151			S82S151			UNIT		
		Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max			
V <sub>IL</sub>	Input voltage <sup>3</sup> Low	V <sub>CC</sub> = Min V <sub>CC</sub> = Max V <sub>CC</sub> = Min, I <sub>in</sub> = -12mA	2.0	.80	2.0	.80	.80	V		
V <sub>IH</sub>	High									
V <sub>IC</sub>	Clamp <sup>3,4</sup>									
V <sub>OL</sub>	Output voltage <sup>3</sup> Low	V <sub>CC</sub> = Min I <sub>OL</sub> = 10mA I <sub>OL</sub> = 8mA I <sub>OH</sub> = -2mA	2.4	.5	2.4	.5	.5	V		
V <sub>OL</sub>	Low									
V <sub>OH</sub>	High									
I <sub>IL</sub>	Input Current Low	V <sub>CC</sub> = Max V <sub>IN</sub> = 0.45V V <sub>IN</sub> = 5.5V		-100 40		-150 50	-150 50	μA		
I <sub>IH</sub>	High									
I <sub>O(OFF)</sub>	Output Current Hi-Z state <sup>7</sup>	V <sub>CC</sub> = max V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = .45V V <sub>OUT</sub> = 0V	-15	80 -140 -70	-15	110 -210 -85	110 -210 -85	μA mA		
I <sub>OS</sub>	Short circuit <sup>4,5</sup>									
I <sub>CC</sub>	V <sub>CC</sub> supply current	V <sub>CC</sub> = max	130	155	130	155	130	155	mA	
C <sub>IN</sub>	Capacitance Input	V <sub>CC</sub> = 5V V <sub>IN</sub> = 2.0V V <sub>B</sub> = 2.0V	8	15		8	15	8	15	pF
C <sub>B</sub>	I/O									

### AC ELECTRICAL CHARACTERISTICS

R<sub>1</sub> = 470Ω, R<sub>2</sub> = 1KΩ  
N82S151: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S151: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TO	FROM	TEST CONDITIONS	N82S151			S82S151			UNIT
				Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max	
T <sub>PD</sub>	Propagation delay	Output ±	Input ±	C <sub>L</sub> = 30pF	20	25	20	20	ns	
T <sub>OE</sub>	Output enable	Output-	Input ±		20	25	20	20	ns	
T <sub>OD</sub>	Output disable <sup>6</sup>	Output+	Input ±	C <sub>L</sub> = 5pF	20	25	20	20	ns	

NOTES

- Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device of these or any other condition above those indicated in the operation of the device specifications is not implied.
- All typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.
- All voltage values are with respect to network ground terminal.
- Test one at a time.
- Duration of short circuit should not exceed 1 second.
- Measured at V<sub>T</sub> = V<sub>OL</sub> + 0.5V.
- I<sub>O(OFF)</sub> at V<sub>OUT</sub> = 0.45V includes I<sub>IL</sub> max value and at V<sub>OUT</sub> = 5.5V includes I<sub>IH</sub> max value.

# FIELD PROGRAMMABLE LOGIC ARRAY (18x42x10) 82S153/82S153A (T.S.)

INTEGRATED FUSE LOGIC  
SERIES 20

## DESCRIPTION

The 82S153 and 82S153A are two-level logic elements, consisting of 32 AND gates and 10 OR gates with fusible link connections for programming I/O polarity and direction.

All AND gates are linked to 8 inputs (I) and 10 bidirectional I/O lines (B). These yield variable I/O gate configurations via 10 direction control gates (D), ranging from 18 inputs to 10 outputs.

On chip T/C buffers couple either True (I, B) or Complement ( $\bar{I}$ ,  $\bar{B}$ ) input polarities to all AND gates, whose outputs can be optionally linked to all OR gates. Their output polarity, in turn, is individually programmable through a set of EX-OR gates for implementing AND/OR or AND/NOR logic functions.

The 82S153 and the 82S153A are field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

Both devices are available in a 20-pin slim line package. For the commercial temperature range (0°C to +75°C) specify N82S153 N or F and N82S153A N or F. For the military temperature range (-55°C to +125°C) specify S82S153 F only and S82S153A F only.

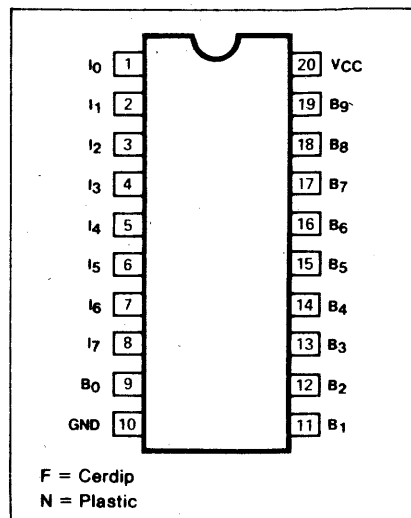
## FEATURES

- Field Programmable (Ni-Cr links)
- 8 inputs
- 32 AND gates
- 10 OR gates
- 10 bidirectional I/O lines
- Active high or low outputs
- 42 Product Terms:
  - 32 Logic Terms
  - 10 Control Terms
- I/O propagation delay:
  - N82S153: 40ns (max)
  - N82S153A: 30ns (max)
  - S82S153: 60ns (max)
  - S82S153A: 45ns (max)
- Input loading
  - N82S153: -100µA (max)
  - S82S153: -150µA (max)
- Power dissipation: 650mW (typ)
- Output: three-state
- TTL compatible

## APPLICATIONS

- Random logic
- Code converters
- Fault detectors
- Function generators
- Address mapping
- Multiplexing

## PIN CONFIGURATION



## LOGIC FUNCTION

Typical product term:  
 $P_n = A \cdot \bar{B} \cdot C \cdot D \dots$

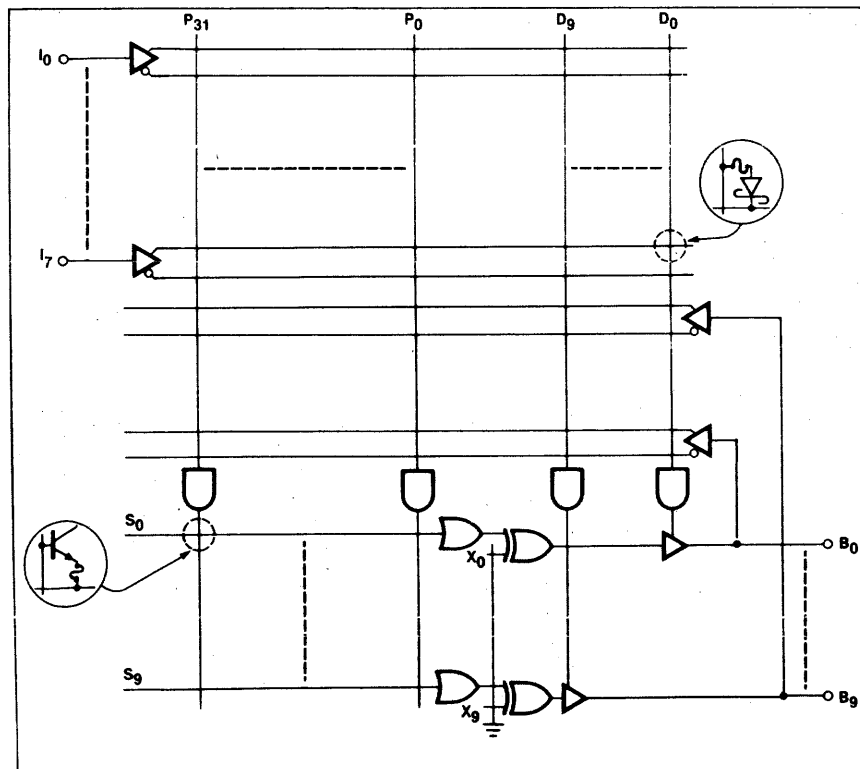
Typical logic function:  
At Output Polarity = H  
 $Z = P_0 + P_1 + P_2 \dots$

At Output Polarity = L  
 $Z = P_0 + P_1 + P_2 \dots$   
 $Z = P_0 \cdot P_1 \cdot P_2 \dots$

### NOTES:

1. For each of the 10 outputs, either function Z (active-high) or  $\bar{Z}$  (active-low) is available, but not both. The desired output polarity is programmed via the EX-OR gates.
2. Z, A, B, C, etc. are user defined connections to fixed inputs (I) and bidirectional pins (B).

## FUNCTIONAL DIAGRAM



# FIELD PROGRAMMABLE LOGIC ARRAY (18x42x10) 82S153/82S153A (T.S.)

INTEGRATED FUSE LOGIC  
SERIES 20

## ABSOLUTE MAXIMUM RATINGS<sup>1</sup>

PARAMETER	RATING		UNIT
	Min	Max	
V <sub>CC</sub>	Supply voltage	+7	Vdc
V <sub>IN</sub>	Input voltage	+5.5	Vdc
V <sub>OUT</sub>	Output voltage	+5.5	Vdc
I <sub>IN</sub>	Input currents	-30	mA
I <sub>OUT</sub>	Output currents	+100	mA
T <sub>A</sub>	Temperature range		C°
	Operating		
	N82S153/153A	0	+75
	S82S153/153A	-55	+125
T <sub>STG</sub>	Storage	-65	+150

## THERMAL RATINGS

TEMPERATURE	Military	Commercial
Maximum junction	175°C	150°C
Maximum ambient	125°C	75°C
Allowable thermal rise ambient to junction	50°C	75°C

## DC ELECTRICAL CHARACTERISTICS

N82S153, N82S153A: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75 ≤ V<sub>CC</sub> ≤ 5.25V

S82S153, S82S153A: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TEST CONDITION	N82S153			S82S153			UNIT
		Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max	
V <sub>IL</sub>	Input voltage <sup>3</sup> Low			.85			.80	V
V <sub>IH</sub>	High	2.0		-1.2	2.0		-1.2	
V <sub>IC</sub>	Clamp <sup>3,4</sup>		- .8			- .8		
V <sub>OL</sub>	Output voltage Low <sup>3,5</sup>			.5			.5	V
V <sub>OL</sub>	Low <sup>3,5</sup>							
V <sub>OH</sub>	High <sup>3,6</sup>	2.4			2.4			
I <sub>IL</sub>	Input current Low			-100			-150	μA
I <sub>IH</sub>	High			40			50	
I <sub>O(OFF)</sub>	Output current Hi-Z state <sup>10</sup>			80			110	μA
I <sub>OS</sub>	Short circuit <sup>4, 6, 7</sup>	-15		-140	-15		-210	mA
I <sub>CC</sub>	V <sub>CC</sub> supply current <sup>8</sup>		130	155		130	165	mA
C <sub>IN</sub>	Capacitance Input		8			8		pF
C <sub>B</sub>	I/O		15			15		

## AC ELECTRICAL CHARACTERISTICS

R<sub>A</sub> = 470Ω, R<sub>2</sub> = 1KΩ

N82S153, N82S153A: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75 ≤ V<sub>CC</sub> ≤ 5.25V

S82S153, S82S153A: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TO	FROM	TEST CONDITIONS	N82S153			S82S153			UNIT
				Min	Typ	Max	Min	Typ	Max	
T <sub>PD</sub>	Propagation delay	Output ±	Input ±		30	40		30	55	ns
T <sub>OE</sub>	Output enable	Output-	Input ±		25	35		25	45	ns
T <sub>OD</sub>	Output disable <sup>9</sup>	Output+	Input ±		25	35		25	45	ns

PARAMETER	TO	FROM	TEST CONDITIONS	N82S153A			S82S153A			UNIT
				Min	Typ	Max	Min	Typ	Max	
T <sub>PD</sub>	Propagation delay	Output ±	Input ±		20	30		20	45	ns
T <sub>OE</sub>	Output enable	Output-	Input ±		20	30		20	40	ns
T <sub>OD</sub>	Output disable <sup>9</sup>	Output+	Input ±		20	30		20	40	ns

**FIELD PROGRAMMABLE LOGIC SEQUENCER  
(16x45x12)**

**82S159 (T.S.)**

INTEGRATED FUSE LOGIC  
SERIES 20

**DESCRIPTION**

The 82S159 is a three state output registered logic element combining AND/OR gate arrays with clocked J/K flip-flops, dynamically convertible to D-type via a "foldback" inverting buffer and control gate F<sub>C</sub>. It features 8 registered I/O outputs (F), in conjunction with bidirectional I/O lines (B). These yield variable I/O gate and register configurations via control gates (D, L) ranging from 16 inputs to 12 outputs.

The AND/OR arrays consist of 32 logic AND gates, 13 control AND gates, and 21 OR gates with fusible link connections for programming I/O polarity and direction. All AND gates are linked to 4 inputs (I), bidirectional I/O lines (B), internal flip-flop outputs (Q), and Complement Array output (C̄). The Complement Array consists of a NOR gate optionally linked to all AND gates for generating and propagating complementary AND terms.

On chip I/O buffers couple either True (I, B, Q) or Complement (Ī, B̄, Q̄, C̄) input polarities to all AND gates, whose outputs can be optionally linked to all OR gates. One group of OR gates drives bidirectional I/O lines (B), whose output polarity is individually programmable through a set of EX-OR gates for implementing AND-OR or AND-NOR logic functions. Another group drives the J-K inputs of all flip-flops, as well as asynchronous Preset and Reset lines (P, R), (except the 82S158/159, where AND functions are provided).

All flip-flops are positive edge trigger and can be used as input, output, or I/O (for interfacing with a bidirectional data bus) in conjunction with load control gates (L), steering inputs (I), (B), (Q) and programmable output select lines (E).

The 82S159 is field programmable, enabling the user to quickly generate custom patterns using standard programming equipment.

The 82S159 is available in a 20-pin, slim line package. For the commercial temperature range (0°C to +75°C) specify N82S159N or F. For the military range (-55°C to +125°C) specify S82S159F only.

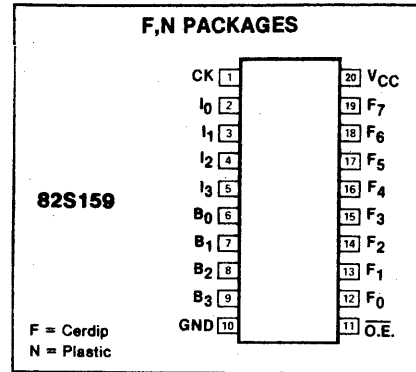
**FEATURES**

- Field programmable (Ni-Cr link)
- 4 dedicated inputs
- 13 control gates
- 32 AND gates
- 21 OR gates
- 45 product terms:  
32 logic terms  
13 control terms
- 4 bidirectional I/O lines
- 8 bidirectional registers
- J/K, T, or D-type flip-flops
- Asynchronous Preset/Reset
- Complement Array
- Active high or low outputs
- Programmable O.E. control
- Positive edge trigger clock
- Power-on reset on all flip-flops (F<sub>n</sub> = "1")
- Clock frequency: N82S159: 15 MHz (max)  
S82S159:        MHz (max)
- Input loading: N82S159: -100µA (max)  
S82S159: -150µA (max)
- Power dissipation: 650mW (typ)
- TTL Compatible

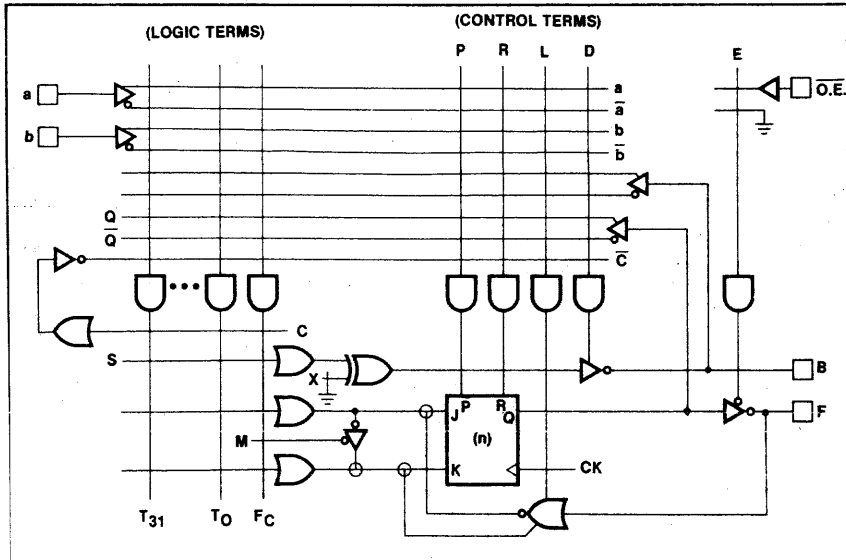
**APPLICATIONS**

- Random sequential logic
- Synchronous up/down counters
- Shift registers
- Bidirectional data buffers
- Timing function generators
- System controllers/synchronizers
- Priority encoder/registers

**PIN CONFIGURATION**



**FUNCTIONAL DIAGRAM**



**FLIP-FLOP TRUTH TABLE**

VCC	O.E.	L	CK	P	R	J	K	Q	F
	H								H/H/Z
	↑	L	X	X	L	X	X	L	H
+5	L	X	X	H	L	X	X	H	L
	L	X	X	L	H	X	X	L	H
	L	L	↑	L	L	L	L	Q	Q̄
	L	L	↑	L	L	L	H	L	H
	L	L	↑	L	L	H	L	H	L
	L	L	↑	L	L	H	H	Q̄	Q
H	H	↑	L	L	L	H	L	H*	
H	H	↑	L	L	H	L	H	L*	
+10V	X	↑	X	X	L	H	L	H**	
	X	↑	X	X	H	L	H	L**	

**NOTES**

1. Positive Logic:  
J/K = T<sub>0</sub> + T<sub>1</sub> + T<sub>2</sub> + ... + T<sub>47</sub>  
T<sub>n</sub> = C̄ • (I<sub>0</sub> • I<sub>1</sub> • I<sub>2</sub> • ...) • (Q<sub>0</sub> • Q<sub>1</sub> • ...) • (B<sub>0</sub> • B<sub>1</sub> • ...)
2. ↑ denotes transition from Low to High level.
3. X = Don't Care
4. \* = Forced at F<sub>n</sub> pin for loading J/K flip-flop in I/O mode. L must be enabled, and other active T<sub>n</sub> disabled via steering input(s) I, B, or Q.
5. At P = R = H, Q = H. The final state of Q depends on which is released first.
6. \*\* = Forced at F<sub>n</sub> pin to load J/K flip-flop independent of program code (Diagnostic mode).



**FIELD PROGRAMMABLE LOGIC SEQUENCER  
(16x45x12)**

**82S159 (T.S.)**

INTEGRATED FUSE LOGIC  
SERIES 20

**ABSOLUTE MAXIMUM RATINGS<sup>1</sup>**

PARAMETER	RATING		UNIT
	Min	Max	
V <sub>CC</sub>	Supply voltage	+7	Vdc
V <sub>IN</sub>	Input voltage	+5.5	Vdc
V <sub>OUT</sub>	Output voltage	+5.5	Vdc
I <sub>IN</sub>	Input currents	+30	mA
I <sub>OUT</sub>	Output currents	+100	mA
T <sub>A</sub>	Temperature range		°C
T <sub>A</sub>	Operating	N82S159	+75
		S82S159	-55
T <sub>STG</sub>	Storage	-65	+150

**THERMAL RATINGS**

TEMPERATURE	MILI-TARY	COM-MER-CIAL
Maximum junction	175°C	150°C
Maximum ambient	125°C	75°C
Allowable thermal rise ambient to junction	50°C	75°C

**DC ELECTRICAL CHARACTERISTICS** N82S159: 0° ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S159: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TEST CONDITIONS	N82S159			S82S159			UNIT	
		Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max		
V <sub>IH</sub> V <sub>IL</sub> V <sub>IC</sub>	Input voltage <sup>3</sup> High Low Clamp	V <sub>CC</sub> = Max V <sub>CC</sub> = Min V <sub>CC</sub> = Min, I <sub>IN</sub> = -12mA	2		0.80	2		0.8	V
V <sub>OH</sub> V <sub>OL</sub> V <sub>OL</sub>	Output voltage <sup>3</sup> High (82S155/7/9) Low Low	V <sub>CC</sub> = Min I <sub>OH</sub> = -2mA I <sub>OL</sub> = 10mA I <sub>OL</sub> = 10mA	2.4	0.35	0.5	2.4	0.35	0.5	V
I <sub>IH</sub> I <sub>IL</sub> I <sub>IL</sub>	Input current High Low Low (CK input)	V <sub>CC</sub> = Max V <sub>IN</sub> = 5.5V V <sub>IN</sub> = 0.45V V <sub>IN</sub> = 0.45V		<1 -10 -50	40 -100 -250		<1 -10 -50	50 -150 -350	μA
I <sub>O(OFF)</sub> I <sub>OS</sub>	Output current Hi-Z state <sup>8</sup> Short circuit <sup>4,6</sup>	V <sub>CC</sub> = Max V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = 0.45V V <sub>OUT</sub> = 0V		1 -1	80 -140		1 -1	110 -210	μA mA
I <sub>CC</sub>	V <sub>CC</sub> supply current <sup>7</sup>	V <sub>CC</sub> = Max		150	190		150	190	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output	V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V V <sub>OUT</sub> = 2.0V		8 15			8 15		pF

**NOTES**

- Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device of these or any other condition above those indicated in the operation of the device specifications is not implied.
- All typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.
- All voltage values are with respect to network ground terminal.
- Test one at a time.
- Measured with V<sub>IH</sub> applied to  $\overline{O.E.}$ .
- Duration of short circuit should not exceed 1 second.
- I<sub>CC</sub> is measured with the  $\overline{O.E.}$  input grounded, all other inputs at 4.5V and the outputs open.
- I<sub>O(OFF)</sub> at V<sub>OUT</sub> = 0.45V includes I<sub>IL</sub> max value and at V<sub>OUT</sub> = 5.5V includes I<sub>IH</sub> max value.

**FIELD PROGRAMMABLE LOGIC SEQUENCER  
(16x45x12)**

**82S159 (T.S.)**

INTEGRATED FUSE LOGIC  
SERIES 20

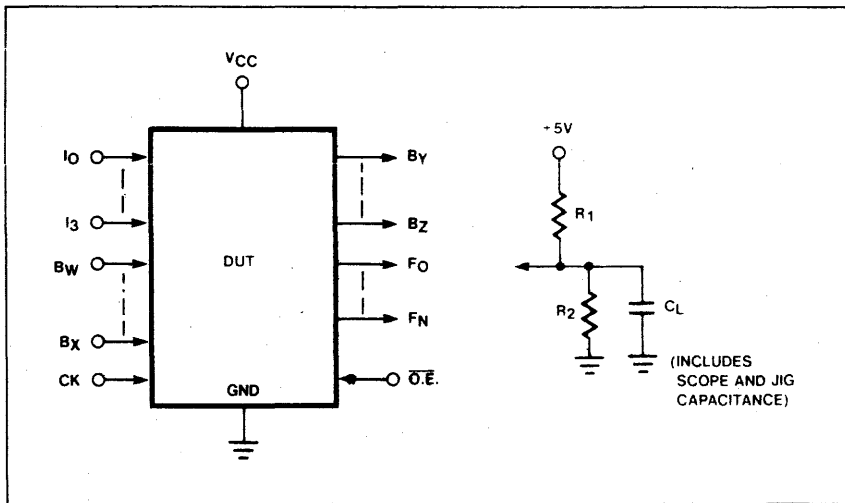
**AC ELECTRICAL CHARACTERISTICS.**  $R_1 = 470\Omega, R_2 = 1k\Omega$

N82S159:  $0^\circ\text{C} \leq T_A \leq +75^\circ\text{C}, 4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S159:  $-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}, 4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

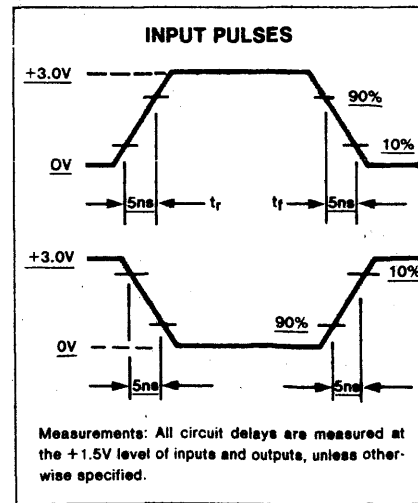
PARAMETER	TO	FROM	TEST CONDITIONS	N82S159			S82S159			UNIT
				Min <sup>5</sup>	Typ <sup>1</sup>	Max	Min	Typ <sup>1</sup>	Max	
T <sub>CKH</sub> Pulse width Clock <sup>2</sup> high T <sub>CKL</sub> Clock low T <sub>CKP</sub> Period T <sub>PRH</sub> Preset/Reset pulse	CK-	CK+	C <sub>L</sub> = 30pF	25	20			20		ns
				30	20			20		
				70	50			50		
				40	30			20		
T <sub>IS1</sub> Set up time Input T <sub>IS2</sub> Input (through F <sub>n</sub> ) T <sub>IS3</sub> Input (through Complement array) <sup>4</sup>	CK+	(I,B)±	C <sub>L</sub> = 30pF	40	30			30	ns	
				—	10			5		
				65	40			40		
T <sub>IH1</sub> Hold time Input T <sub>IH2</sub>	CK+	(I,B)±	C <sub>L</sub> = 30pF		-10	0		-10	ns	
					0	—		-5		
T <sub>CKO</sub> Propagation delay Clock T <sub>OE1</sub> Output enable T <sub>OD1</sub> Output disable <sup>3</sup> T <sub>PD</sub> Output T <sub>OE2</sub> Output enable T <sub>OD2</sub> Output disable <sup>3</sup> T <sub>PRO</sub> Preset/Reset T <sub>PPR</sub> Power-on preset	F±	CK+	C <sub>L</sub> = 30pF		25	30		25	ns	
					20	30		20		
					20	55		20		
					40	50		35		
					35	55		35		
					35	35		35		
					50	55		50		
	0	10		0						

- NOTE
- All typical values are at  $V_{CC} = 5\text{V}, T_A = 25^\circ\text{C}$ .
  - To prevent spurious clocking, clock rise time (10%-90%) < 10ns.
  - Measured at  $V_T = V_{OL} + 0.5\text{V}$ .
  - When using the Complement Array  $T_{CKP} = 75\text{ns}$  (min).
  - Limits are guaranteed with 12 product terms maximum connected to each sum term line.

**TEST LOAD CIRCUIT**



**VOLTAGE WAVEFORM**



MEMORY

# FIELD PROGRAMMABLE LOGIC SEQUENCER (14x48x6)

82S167 (T.S.)

INTEGRATED FUSE LOGIC  
SERIES 24

## DESCRIPTION

The 82S167 is a bipolar, programmable state machine of the Mealy type. It contains logic AND-OR gate arrays with user programmable connections which control the inputs of on-chip state and output registers. These consist respectively of 8  $Q_p$  and 4  $Q_f$  edge-triggered, clocked S/R flip-flops, with an asynchronous preset option.

All flip-flops are unconditionally preset to "1" during power turn-on.

The AND array combines 14 external inputs,  $I_{0-13}$ , with 8 internal inputs,  $P_{0-7}$ , fed back from the State register to form up to 48 transition terms (AND terms). In addition, bit-0 and bit-1 of the internal state register are brought off-chip to allow extending the output register to 6 bits, if so desired.

All transition terms can include True, False, or Don't Care states of the controlling variables, and are merged in the OR array to issue next-state and next-output commands to their respective registers on the low to high transition of the Clock pulse.

Both True and Complement transition terms can be generated by optional use of the internal variable (C) from the complement array. Also, if desired, the Preset in-

## FEATURES

- Field programmable (Ni-Cr link)
- 14 True/Complement buffered inputs
- 48 programmable AND gates
- 25 programmable OR gates
- 8-bit state register
- 2-bit shared state/output register
- 4-bit output register
- Transition complement array
- Programmable asynchronous preset/output enable
- Positive edge-trigger clock
- Power-on preset to logic "1" of all registers
- Automatic logic "HOLD" state via S/R flip-flops
- On-chip test array
- $f_{(max)}$ : 13.9MHz (82S167)
- Power: 650mW (typ)
- TTL compatible
- Tri-state outputs
- Single +5V supply
- 300 mil wide 24-pin DIP

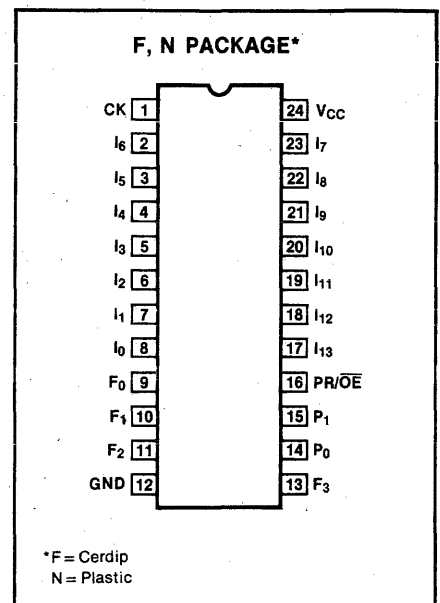
put can be converted to output-enable function, as an additional user programmable option.

The device is available in the commercial temperature range. For the commercial temperature range (0°C to +75°C) specify N82S167F or N.

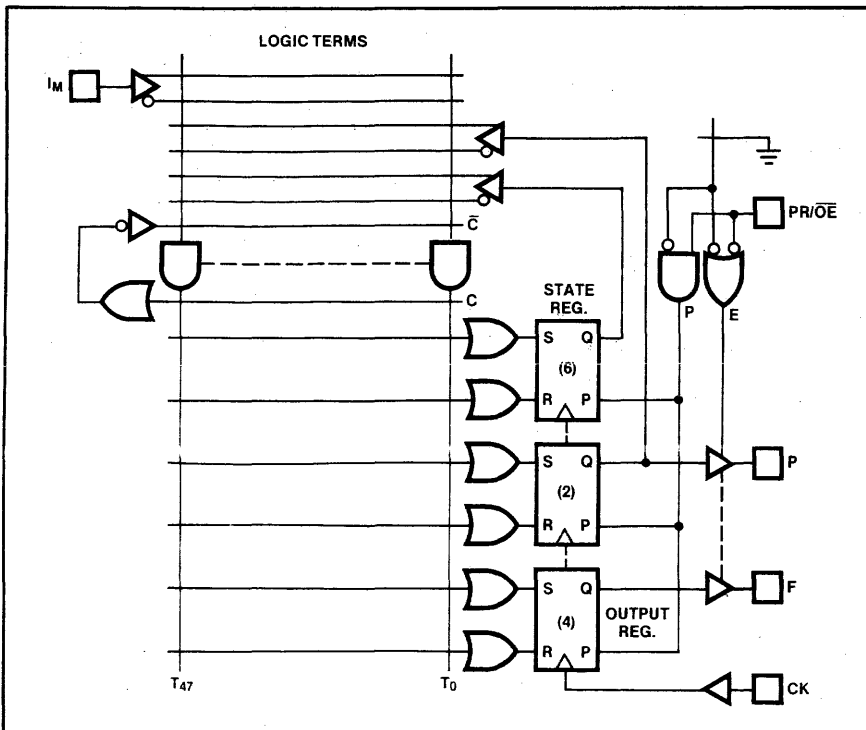
## APPLICATIONS

- Interface protocols
- Sequence detectors
- Peripheral controllers
- Timing generators
- Sequential circuits
- Elevator controllers
- Security locking systems
- Counters
- Shift registers

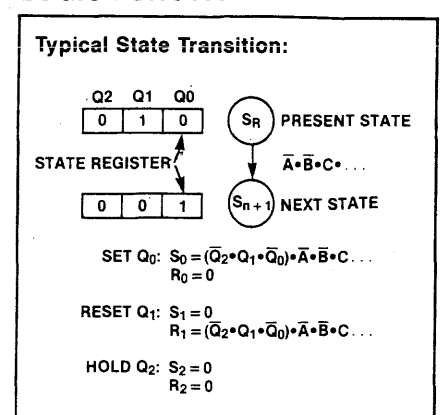
## PIN CONFIGURATION



## LOGIC DIAGRAM



## LOGIC FUNCTION

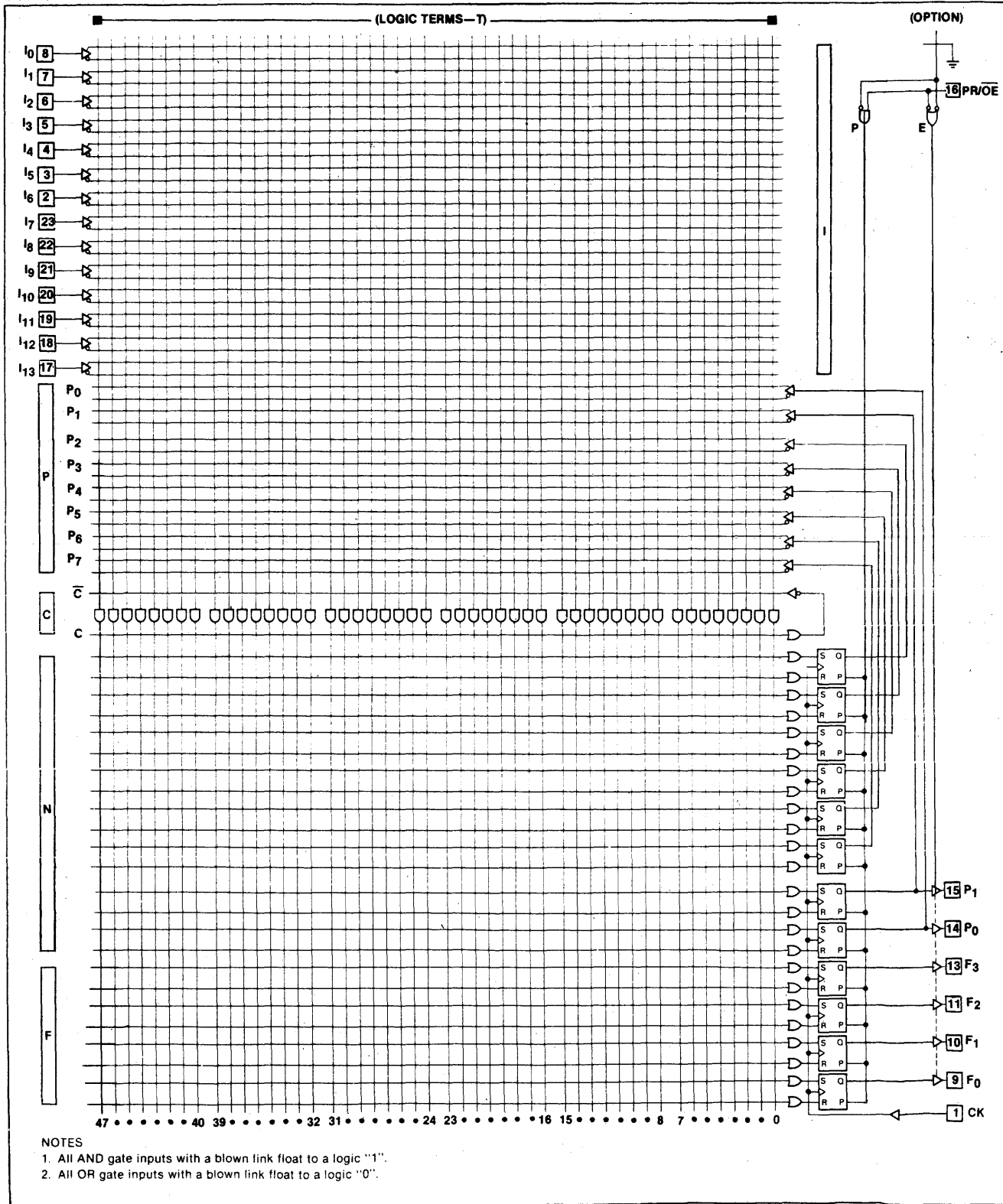


FIELD PROGRAMMABLE LOGIC SEQUENCER (14x48x6)

82S167 (T.S.)

INTEGRATED FUSE LOGIC SERIES 24

FPLS LOGIC DIAGRAM



MEMORY Signetics

# FIELD PROGRAMMABLE LOGIC ARRAY (16x48x8) 82S100 (T.S.)/82S101 (O.C.)

INTEGRATED FUSE LOGIC  
SERIES 28

## DESCRIPTION

The 82S100 (tri-state) and 82S101 (open collector) are Bipolar, Fuse-Link Programmable Logic Arrays (FPLA). Each device utilizes the standard AND/OR/Invert architecture to directly implement custom sum of product logic equations.

Each device consists of 16 dedicated inputs and 8 dedicated outputs. Each output is capable of being actively controlled by any or all of the 48 product terms. The true, complement, or don't care condition of each of the 16 inputs ANDed together comprise one P-term. All 48 P-terms are then ORed to each output. The user must then only select which P-terms will activate an output by disconnecting terms which do not affect the output. In addition each output can be fused as active-high<sup>(H)</sup> or active-low<sup>(L)</sup>.

The 82S100 and 82S101 are fully TTL compatible, and include chip-enable control for expansion of input variables, and output inhibit. They feature either open collector or tri-state outputs for ease of expansion of product terms and application in bus-organized systems.

Both devices are available in commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S100/101, F or N, and for the military temperature range (-55°C to +125°C) specify S82S100/101, F, G or R.

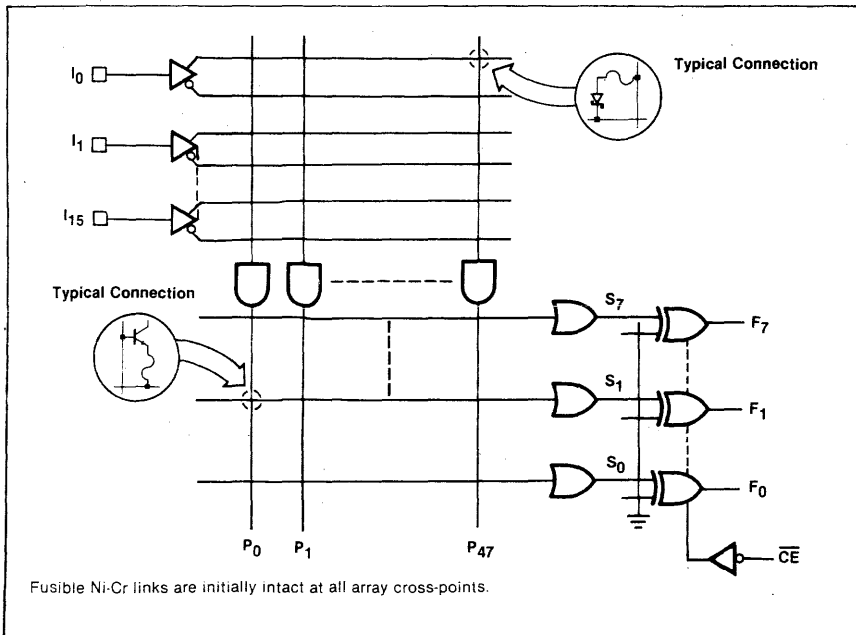
## FEATURES

- Field programmable (Ni-Cr link)
- Input variables: 16
- Output functions: 8
- Product terms: 48
- Address access time:  
S82S100/101 — 80ns max  
N82S100/101 — 50ns max
- Power dissipation: 600mW typ
- Input loading:  
S82S100/101: -150µA max  
N82S100/101: -100µA max
- Chip enable input
- Output option:  
82S100: Tri-state  
82S101: Open collector
- Output disable function:  
Tri-state — Hi-Z  
Open collector — Hi
- Separate I/O architecture

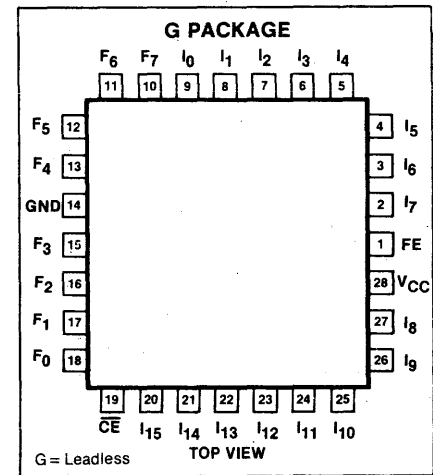
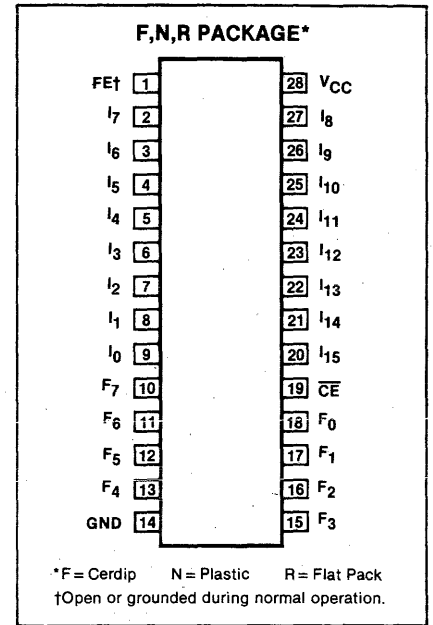
## APPLICATIONS

- CRT display systems
- Random logic
- Code conversion
- Peripheral controllers
- Function generators
- Look-up and decision tables
- Microprogramming
- Address mapping
- Character generators
- Data security encoders
- Fault detectors
- Frequency synthesizers
- 16 bit-to-8 bit bus interface
- Random logic replacement

## LOGIC DIAGRAM



## PIN CONFIGURATION



# FIELD PROGRAMMABLE LOGIC ARRAY (16x48x8) 82S100 (T.S.)/82S101 (O.C.)

INTEGRATED FUSE LOGIC  
SERIES 28

## ABSOLUTE MAXIMUM RATINGS<sup>1</sup>

PARAMETER	RATING		UNIT
	Min	Max	
V <sub>CC</sub> Supply voltage		+ 7	Vdc
V <sub>IN</sub> Input voltage		+ 5.5	Vdc
V <sub>OUT</sub> Output voltage		+ 5.5	Vdc
I <sub>IN</sub> Input currents	- 30	+ 30	mA
I <sub>OUT</sub> Output currents		+ 100	mA
T <sub>A</sub> Temperature range Operating			°C
	N82S100/101	0 + 75	
	S82S100/101	- 55 + 125	
T <sub>STG</sub> Storage	- 65	+ 150	

## THERMAL RATINGS

TEMPERATURE	MILITARY	COMMERCIAL
Maximum junction	175°C	150°C
Maximum ambient	125°C	75°C
Allowable thermal rise ambient to junction	50°C	75°C

## DC ELECTRICAL CHARACTERISTICS

N82S100/101: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S100/101: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TEST CONDITIONS	N82S100/101			S82S100/101			UNIT
		Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max	
V <sub>IH</sub> Input voltage <sup>3</sup> High	V <sub>CC</sub> = Max V <sub>CC</sub> = Min V <sub>CC</sub> = Min, I <sub>IN</sub> = -12mA	2			2			V
V <sub>IL</sub> Low				0.8			0.8	
V <sub>IC</sub> Clamp <sup>3,4</sup>		- 0.8		- 1.2	- 0.8		- 1.2	
V <sub>OH</sub> Output voltage High (82S100) <sup>3,5</sup>	V <sub>CC</sub> = Min I <sub>OH</sub> = -2mA I <sub>OL</sub> = 9.6mA	2.4			2.4			V
V <sub>OL</sub> Low <sup>3,6</sup>			0.35	0.45		0.35	0.50	
I <sub>IH</sub> Input current High	V <sub>IN</sub> = 5.5V V <sub>IN</sub> = 0.45V		< 1	25		< 1	50	μA
I <sub>IL</sub> Low			- 10	- 100		- 10	- 150	
I <sub>OLK</sub> Output current Leakage <sup>7</sup>	CE = High, V <sub>CC</sub> = Max V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = 0.45V CE = Low, V <sub>OUT</sub> = 0V		1	40		1	60	μA
I <sub>O(OFF)</sub> Hi-Z state (82S100) <sup>7</sup>			1	40		1	60	μA
I <sub>OS</sub> Short circuit (82S100) <sup>4,8</sup>			- 15	- 1	- 40	- 15	- 1	- 60
I <sub>CC</sub> V <sub>CC</sub> supply current <sup>9</sup>	V <sub>CC</sub> = Max		120	170		120	180	mA
C <sub>IN</sub> Capacitance <sup>7</sup> Input	CE = High, V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V V <sub>OUT</sub> = 2.0V		8			8		pF
C <sub>OUT</sub> Output				17			17	

## AC ELECTRICAL CHARACTERISTICS

R<sub>1</sub> = 470Ω, R<sub>2</sub> = 1kΩ, C<sub>L</sub> = 30pF  
N82S100/101: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S100/101: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TO	FROM	N82S100/101			S82S100/101			UNIT
			Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max	
T <sub>1A</sub> Propagation delay Input	Output	Input		35	50		35	80	ns
T <sub>CE</sub> Chip enable			Output	Chip enable		15	30		
T <sub>CD</sub> Disable time Chip disable	Output	Chip enable		15	30		15	50	ns

NOTES:

- Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and only functional operation of the device at these or any other conditions above those indicated in the operation of the device specifications is not implied.
- All voltages are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.
- All voltage values are with respect to network ground terminal.
- Test one at a time.
- Measured with V<sub>IL</sub> applied to CE and a logic high stored.
- Measured with a programmed logic condition for which the output test is at a low logic level. Output sink current is applied through a resistor to V<sub>CC</sub>.
- Measured with V<sub>IH</sub> applied to CE.
- Duration of short circuit should not exceed 1 second.
- I<sub>CC</sub> is measured with the chip enable input grounded, all other inputs at 4.5V and the outputs open.

**FIELD PROGRAMMABLE GATE ARRAY (16x9x9)**

**82S102 (O.C.)/82S103 (T.S.)**

INTEGRATED FUSE LOGIC  
SERIES 28

**DESCRIPTION**

The 82S102 and 82S103 are Bipolar, fuse programmable, gate arrays. The device consists of 9 AND/NAND Gates which share 16 common inputs. The type of gate is selected by programming the output as active-high<sup>(H)</sup> or active-low<sup>(L)</sup>. Each of the 16 inputs I<sub>0</sub>-I<sub>15</sub> can be programmed to provide the True (H), Complement (L), or Don't Care (-) state to each of the 9 AND/NAND gates. OR/NOR logic functions can also be implemented by complementing the inputs and outputs via on-chip inverting buffers.

Both devices are field-programmable, which means that custom patterns are immediately available.

The 82S102 and 82S103 include chip-enable control for output strobing and inhibit. They feature either open collector or tri-state outputs for ease of expansion of input variables and application in bus-organized systems.

Both devices are available in the commercial and military temperature ranges. For the commercial range (0°C to +75°C) specify N82S102/103, F or N, and for the military range (-55°C to +125°C) specify S82S102/103, F, G, and R.

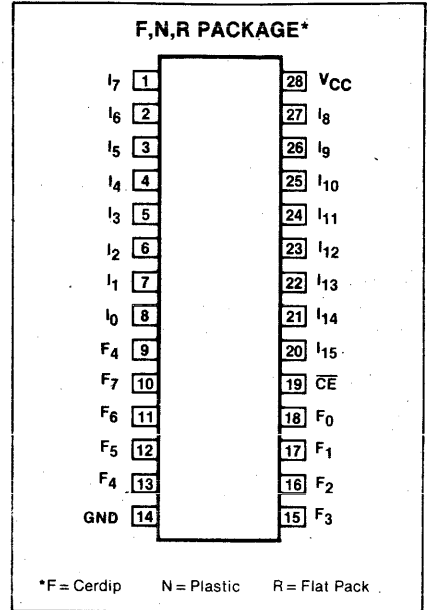
**FEATURES**

- Field programmable (Ni-Cr link)
- 16 input variables
- 9 output functions
- Chip enable input
- I/O propagation delay:  
N82S102/103: 35ns max  
S82S102/103: 50ns max
- Power dissipation: 600mW typ
- Input loading:  
N82S102/103: -100µA max  
S82S102/103: -150µA max
- Output options:  
82S102: Open collector  
82S103: Tri-state
- Output disable function:  
82S102: Hi  
82S103: Hi-Z
- Fully TTL compatible

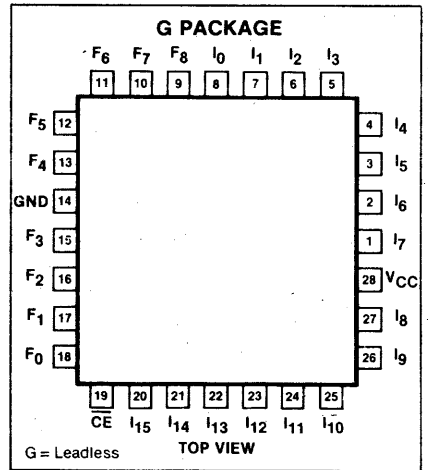
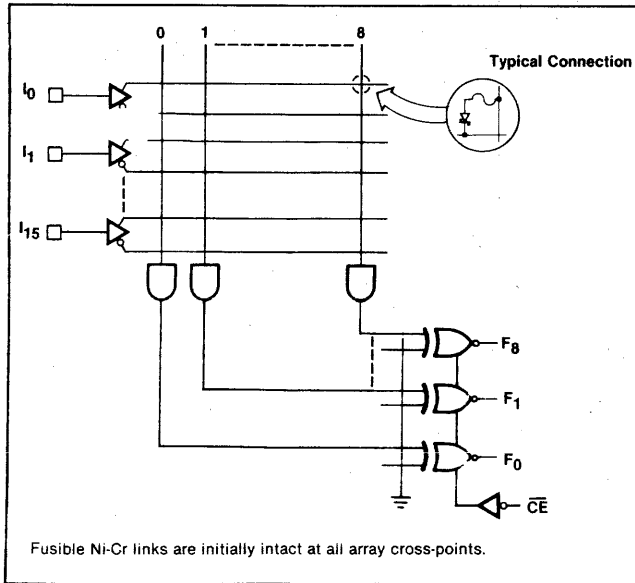
**APPLICATIONS**

- Random logic
- Address decoders
- Code detectors
- Peripheral selectors
- Fault monitors
- Machine state decoders

**PIN CONFIGURATION**



**LOGIC DIAGRAM**



**FIELD PROGRAMMABLE GATE ARRAY (16x9x9) 82S102 (O.C.)/82S103 (T.S.)**

INTEGRATED FUSE LOGIC  
SERIES 28

**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>OH</sub> Output voltage		Vdc
High (82S102)	+5.5	
Off-state (82S103)	+5.5	
I <sub>IN</sub> Input current	±30	mA
I <sub>OUT</sub> Output current	+100	mA
T <sub>A</sub> Temperature range		°C
Operating		
N82S102/103	0 to +75	
S82S102/103	-55 to +125	
T <sub>STG</sub> Storage	-65 to +150	

**THERMAL RATINGS**

TEMPERATURE	MILITARY	COMMERCIAL
Maximum junction	175°C	150°C
Maximum ambient	125°C	75°C
Allowable thermal rise ambient to junction	50°C	75°C

**DC ELECTRICAL CHARACTERISTICS** N82S102/103: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S102/103: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER <sup>1</sup>	TEST CONDITIONS	N82S102/103			S82S102/103			UNIT
		Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max	
V <sub>IL</sub> Input voltage								V
Low <sup>1</sup>	V <sub>CC</sub> = Min	2.0		0.8	2.0		0.8	
High <sup>1</sup>	V <sub>CC</sub> = Max							
V <sub>IC</sub> Clamp <sup>1,3</sup>	V <sub>CC</sub> = Min, I <sub>IN</sub> = -12mA		-0.8	-1.2		-0.8	-1.2	
V <sub>OL</sub> Output voltage								V
Low <sup>1,4</sup>	V <sub>CC</sub> = Min I <sub>OL</sub> = 9.6mA	2.4	0.35	0.45	2.4	0.35	0.50	
V <sub>OH</sub> High (82S103) <sup>1,5</sup>	I <sub>OH</sub> = -2mA							
I <sub>IL</sub> Input current								µA
Low	V <sub>IN</sub> = 0.45V		-10	-100		-10	-150	
High	V <sub>IN</sub> = 5.5V		<1	25		<1	50	
I <sub>OLK</sub> Output current								µA
Leakage (82S102) <sup>6</sup>	V <sub>CC</sub> = Max V <sub>OUT</sub> = 5.5V		1	40		1	60	
I <sub>O(OFF)</sub> Hi-Z state (82S103) <sup>6</sup>	V <sub>OUT</sub> = 5.5V		1	40		1	60	µA
Short circuit (82S103) <sup>3,7</sup>	V <sub>OUT</sub> = 0.45V V <sub>OUT</sub> = 0V	-15	-1	-40	-15	-1	-60	mA
I <sub>CC</sub> V <sub>CC</sub> supply current <sup>8</sup>	V <sub>CC</sub> = Max		120	170		120	180	mA
C <sub>IN</sub> Capacitance								pF
Input	V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V		8			8		
Output <sup>6</sup>	V <sub>OUT</sub> = 2.0V		15			15		

**AC ELECTRICAL CHARACTERISTICS** R<sub>1</sub> = 470Ω, R<sub>2</sub> = 1kΩ, C<sub>L</sub> = 30pF  
N82S102/103: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
S82S102/103: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TO	FROM	N82S102/103			S82S102/103			UNIT
			Min	Typ <sup>2</sup>	Max	Min	Typ <sup>2</sup>	Max	
T <sub>IA</sub> Propagation delay									ns
Input	Output	Input		20	35		20	55	
T <sub>CE</sub> Chip enable	Output	Chip enable		15	30		15	45	
T <sub>CD</sub> Disable time									ns
Chip disable	Output	Chip enable		15	30		15	45	

- NOTES
- All voltage values are with respect to network ground terminal.
  - All typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.
  - Test each output one at a time.
  - Measured with a programmed logic condition for which the output under test is at a low logic level. Output sink current is supplied through a resistor to V<sub>CC</sub>.
  - Measured with V<sub>IL</sub> applied to  $\overline{CE}$  and a logic high at the output.
  - Measured with V<sub>IH</sub> applied to  $\overline{CE}$ .
  - Duration of short circuit should not exceed 1 second.
  - I<sub>CC</sub> is measured with the chip enable input grounded, all other inputs at 4.5V and the outputs open.

Signetics  
MEMORY



**FIELD PROGRAMMABLE LOGIC SEQUENCER (16x48x8)**

**82S105A (T.S.)**

INTEGRATED FUSE LOGIC SERIES 28

**DESCRIPTION**

The 82S105A is a bipolar programmable state machine of the Mealy type. It contains logic AND-OR gate arrays with user programmable connections which control the inputs of on-chip State and Output registers. These consist respectively of 6  $Q_p$ , and 8  $Q_f$  edge triggered, clocked S/R flip-flops, with an asynchronous Preset option. All flip-flops are unconditionally preset to "1" during power turn on.

The AND array combines 16 external inputs  $I_0-15$  with 6 internal inputs  $P_0-5$  fed back from the State register to form up to 48 Transition terms (AND terms). All Transition terms can include True, False, or Don't Care states of the controlling variables, and are merged in the OR array to issue next-state and next-output commands to their respective registers on the Low to High transition of the Clock pulse. Both True and Complement Transition terms can be generated by optional use of the internal input variable (C) from the Complement array. Also, if desired, the Preset input can be converted to Output-Enable function, as an additional user programmable option.

The device is available in commercial temperature range. For the commercial temperature range (0°C to +75°) specify N82S105A, F or N.

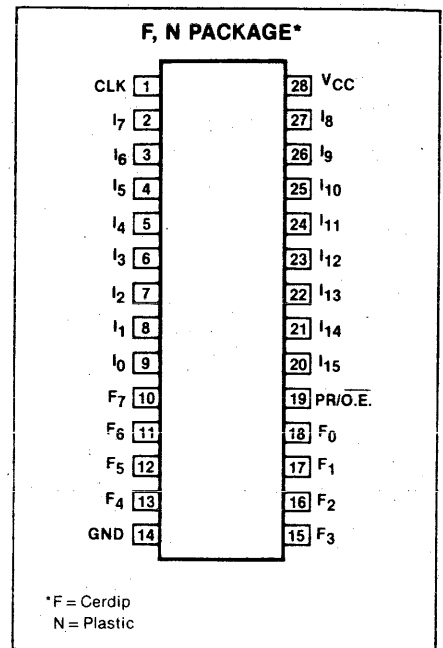
**FEATURES**

- Field programmable (Ni-Cr link)
- 16 input variables
- 8 output functions
- 48 transition terms
- 6-bit state register
- 8-bit output register
- Transition complement array
- Positive edge trigger clock
- Programmable asynchronous preset or output enable
- Power-on preset to all "1" of internal registers
- $f(\text{max}) = 20\text{MHz}$
- 650mW power dissipation (typical)
- TTL compatible
- Single +5V supply
- Open collector and tri-state versions
- TRI-STATE outputs

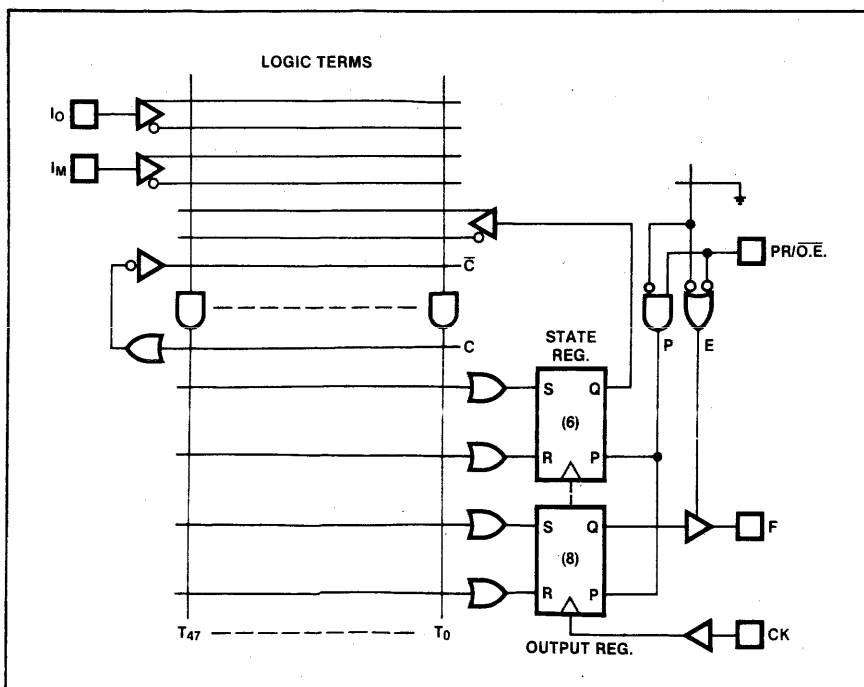
**APPLICATIONS**

- Interface protocols
- Sequence detectors
- Peripheral controllers
- Timing generators
- Sequential circuits
- Elevator controllers
- Security locking systems
- Counters
- Shift registers

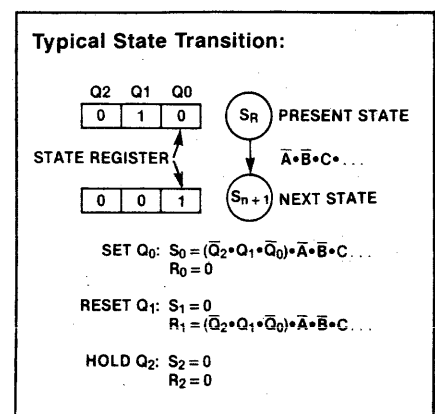
**PIN CONFIGURATION**



**LOGIC DIAGRAM**



**LOGIC FUNCTION**



**FIELD PROGRAMMABLE  
LOGIC SEQUENCER (16 × 48 × 8)**

**82S105A (T.S.)**

INTEGRATED FUSE LOGIC  
SERIES 28

**ABSOLUTE MAXIMUM RATINGS<sup>1</sup>**

PARAMETER	RATING		UNIT
	Min	Max	
V <sub>CC</sub>		+ 7	Vdc
V <sub>IN</sub>		+ 5.5	Vdc
V <sub>OUT</sub>		+ 5.5	Vdc
I <sub>IN</sub>	- 30	+ 30	mA
I <sub>OUT</sub>		+ 100	mA
T <sub>A</sub>		+ 75	°C
T <sub>STG</sub>	- 65	+ 150	
T <sub>SDO</sub>		4	

**THERMAL RATINGS**

TEMPERATURE	MILI-TARY	COM-MER-CIAL
Maximum junction	175°C	150°C
Maximum ambient	125°C	75°C
Allowable thermal rise ambient to junction	50°C	75°C

**DC ELECTRICAL CHARACTERISTICS** 0°C ≤ T<sub>A</sub> ≤ + 75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V

PARAMETER	TEST CONDITIONS	N82S105A			UNIT
		Min	Typ <sup>2</sup>	Max	
V <sub>IH</sub> V <sub>IL</sub> V <sub>IC</sub>	Input voltage <sup>3</sup> High Low Clamp <sup>4</sup> V <sub>CC</sub> = Max V <sub>CC</sub> = Min V <sub>CC</sub> = Min, I <sub>IN</sub> = - 12mA	2		0.8 - 1.2	V
V <sub>OH</sub> V <sub>OL</sub>	Output voltage <sup>3</sup> High <sup>5</sup> Low <sup>6</sup> V <sub>CC</sub> = Min I <sub>OH</sub> = - 2mA I <sub>OL</sub> = 9.6mA	2.4	0.35	0.45	V
I <sub>IH</sub> I <sub>IL</sub> I <sub>IL</sub>	Input current High Low Low (CK input) V <sub>IN</sub> = 5.5V V <sub>IN</sub> = 0.45V V <sub>IN</sub> = 0.45V		< 1 - 10 - 50	25 - 100 - 250	μA
I <sub>O(OFF)</sub> I <sub>OS</sub>	Output current Hi-Z state <sup>7</sup> Short circuit <sup>4,8</sup> V <sub>CC</sub> = Max V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = 0.45V V <sub>OUT</sub> = 0V		1 - 1	40 - 40 - 70	μA mA
I <sub>CC</sub>	V <sub>CC</sub> supply current <sup>9</sup> V <sub>CC</sub> = Max		120	180	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance <sup>7</sup> Input Output V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V V <sub>OUT</sub> = 2.0V		8 10		pF

- NOTES
- Stresses above those listed under "Absolute Maximum Ratings" may cause malfunction or permanent damage to the device. This is a stress rating only. Functional operation at these or any other condition above those indicated in the operational and programming specification of the device is not implied.
  - All typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.
  - All voltage values are with respect to network ground terminal.
  - Test one at a time.
  - Measured with V<sub>IL</sub> applied to  $\overline{O.E.}$  and a logic high stored, or with V<sub>IH</sub> applied to PR.
  - Measured with a programmed logic condition for which the output is at a low logic level, and V<sub>IL</sub> applied to PR/ $\overline{O.E.}$ . Output sink current is supplied through a resistor to V<sub>CC</sub>.
  - Measured with V<sub>IH</sub> applied to PR/ $\overline{O.E.}$ .
  - Duration of short circuit should not exceed 1 second.
  - I<sub>CC</sub> is measured with the PR/ $\overline{O.E.}$  input grounded, all other inputs at 4.5V and the outputs open.
  - See OR DRIVE limitations on page 9.

Signetics  
MEMORY

FIELD PROGRAMMABLE  
LOGIC SEQUENCER (16x48x8)

82S105A (T.S.)

AC ELECTRICAL CHARACTERISTICS  $R_1 = 470\Omega$ ,  $R_2 = 1k\Omega$ ,  $C_L = 30pF$   $0^\circ C \leq T_A \leq +75^\circ C$ ,  $4.75V \leq V_{CC} \leq 5.25V$ 

PARAMETERS	TO	FROM	N82S105A			UNIT
			Min	Typ <sup>1</sup>	Max	
Pulse width $T_{CKH}$ Clock high <sup>2</sup> $T_{CKL}$ Clock low $T_{CKP1}$ B Period (w/o c-array) $T_{CKP2}$ B Period (w/c-array) $T_{PRH}$ Preset pulse	CK -	CK +	25	15		ns
	CK +	CK -	25	15		
	CK +	CK +	50	40		
	CK +	CK +	80	50		
	PR +	PR -	25	15		
Set-up time <sup>3</sup> $T_{IS1}$ A Input $T_{IS1}$ B Input $T_{IS2}$ A Input (through Complement array) $T_{IS2}$ B Input (through Complement array) $T_{VS}$ Power-on preset $T_{PRS}$ Preset.	CK +	Input $\pm$	40			ns
	CK +	Input $\pm$	30			
	CK +	Input $\pm$	70			
	CK +	Input $\pm$	60			
	CK -	$V_{CC}$ +	0	-10		
	CK -	PRS -	0	-10		
Hold time $T_{IH}$ Input	Input $\pm$	CK +	5	-10		ns
Propagation delay $T_{CKO}$ Clock $T_{OE}$ Output enable $T_{OD}$ Output disable $T_{PR}$ Preset $T_{PPR}$ Power-on preset	Output $\pm$	CK +		15	20	ns
	Output -	$\overline{O.E.}$ -		20	30	
	Output +	$\overline{O.E.}$ +		20	30	
	Output +	PR +		18	30	
	Output +	$V_{CC}$ +		0	10	
Frequency of operation $f_{MAX}$ B w/o c-array $f_{MAX}$ B w/c-array					20	MHz
					12.5	

## NOTES:

- All typical values are at  $V_{CC} \pm 5V$ ,  $T_A = 25^\circ C$ .
- To prevent spurious clocking, clock rise time (10%-90%)  $\leq 30ns$ .
- See "Speed vs. OR Loading" on page 9.

# 256-BIT BIPOLAR RAM (256 × 1)

# 82LS16 (T.S.)/82LS17 (O.C.)

## DESCRIPTION

The 82LS16 and 82LS17 are Read/Write memory arrays which feature either open collector or 3-state output options for optimization of word expansion in bused organizations. Memory expansion is further enhanced by full on-chip address decoding, 3 chip enable inputs and PNP input transistors which reduce input loading.

During Write operation, the logical state of the output of both devices follows the complement of the data input being written. This feature allows faster execution of Write/Read cycles, enhancing the performance of systems utilizing indirect addressing modes, and/or requiring immediate verification following a Write cycle.

Both devices have fast Read access and Write cycle times, as well as low power requirements and thus are ideally suited in high-speed memory applications such as cache, buffers, scratch pads, writable control stores, where power limitations are of major concern.

Both devices are available in the commercial temperature range (0°C to +75°C) and the military temperature range (-55°C to +125°C). They are specified as: N82LS16F or N or N82LS17F or N for the commercial temperature range, and S82LS16F, G, or W or S82LS17F, G, or W for the military temperature range. Military products are available as fully processed to Mil-Std 883 Level B or Level C; specify either 883B or 883C.

See page 4-9 for Truth Table, Timing Diagrams, Test Circuit and Waveform.

## FEATURES

- Address access time:  
N82LS16/17: 40ns max  
S82LS16/17: 60ns max
- Write cycle time:  
N82LS16/17: 40ns max  
S82LS16/17: 65ns max
- Power dissipation: 0.98mW/bit typ
- Input loading:  
N82LS16/17: -100µA max  
S82LS16/17: -250µA max

- Output follows complement of data input during Write
- On-chip address decoding
- Output option:  
82LS16 3-state  
82LS17 Open collector
- Schottky clamped
- TTL compatible

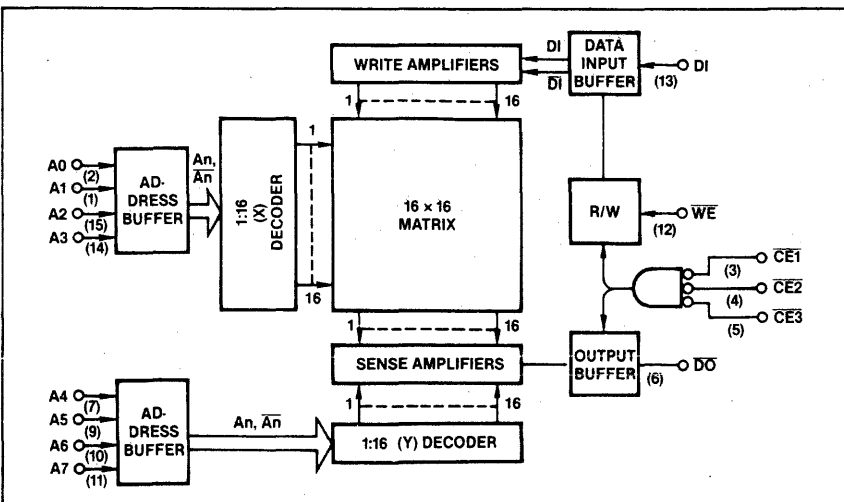
## APPLICATIONS

- Buffer memory
- Writable control store
- Memory mapping
- Push down stack
- Scratch pad

## ABSOLUTE MAXIMUM RATINGS

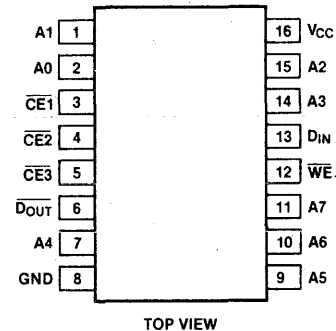
PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>OUT</sub> Output voltage	+5.5	Vdc
T <sub>A</sub> Temperature Range	Operating	°C
	N grade	
	S grade	-55 to +125
T <sub>STG</sub> Storage	-65 to +150	°C

## BLOCK DIAGRAM



## PIN CONFIGURATIONS

### F, N, W PACKAGE

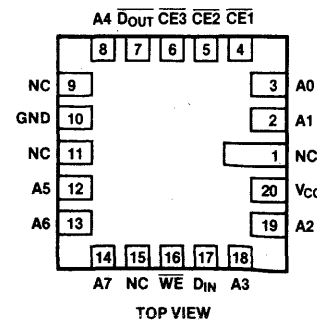


### Order Numbers:

N82LS16F or N82LS16N  
N82LS17F or N82LS17N  
S82LS16F or S82LS16W  
S82LS17F or S82LS17W

F = Cerdip  
N = Plastic  
W = Flat Pak

### G PACKAGE



### Order Numbers:

S82LS16G or S82LS17G

G = Leadless

**256-BIT BIPOLAR RAM (256 × 1)**

**82LS16 (T.S.)/82LS17 (O.C.)**

**DC ELECTRICAL CHARACTERISTICS** N82LS16/17: 0 °C ≤ T<sub>A</sub> ≤ +75 °C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
 S82LS16/17: -55 °C ≤ T<sub>A</sub> ≤ +125 °C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TEST CONDITIONS	N82LS16/17			S82LS16/17			UNIT
		Min	Typ <sup>1</sup>	Max	Min	Typ <sup>1</sup>	Max	
V <sub>IH</sub> V <sub>IL</sub> V <sub>IC</sub>	Input voltage <sup>2</sup> High Low Clamp <sup>3</sup> V <sub>CC</sub> = max V <sub>CC</sub> = min V <sub>CC</sub> = min, I <sub>IN</sub> = -12mA	2.0		0.8 -1.5	2.0		0.8 -1.5	V
V <sub>OH</sub> V <sub>OL</sub>	Output voltage <sup>2</sup> High (82LS16) <sup>4</sup> Low <sup>5</sup> V <sub>CC</sub> = min I <sub>OH</sub> = -3.2mA I <sub>OL</sub> = 16mA	2.6	0.35	0.45	2.4	0.35	0.5	V
I <sub>IH</sub> I <sub>IL</sub>	Input current <sup>3</sup> High Low V <sub>CC</sub> = max V <sub>IN</sub> = 5.5V V <sub>IN</sub> = 0.45V		1 -10	25 -100		1 -10	25 -250	μA
I <sub>OLK</sub> I <sub>OZ</sub> I <sub>OS</sub>	Output current Leakage (82LS17) <sup>5</sup> Hi-Z state (82LS16) <sup>6</sup> Short circuit (82LS16) <sup>7</sup> V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = 5.5V V <sub>OUT</sub> = 0.45V V <sub>CC</sub> = max, V <sub>O</sub> = 0V		1 1 -1	40 40 -40 -70		1 1 -1	40 50 -50 -70	μA
I <sub>CC</sub>	V <sub>CC</sub> supply current <sup>8</sup> V <sub>CC</sub> = max		50	70		50	100	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V V <sub>OUT</sub> = 2.0V		5 8			5 8		pF

**AC ELECTRICAL CHARACTERISTICS** R<sub>1</sub> = 270Ω, R<sub>2</sub> = 600Ω, C<sub>L</sub> = 30pF  
 N82LS16/17: 0 °C ≤ T<sub>A</sub> ≤ +75 °C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V  
 S82LS16/17: -55 °C ≤ T<sub>A</sub> ≤ +125 °C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TO	FROM	N82LS16/17			S82LS16/17			UNIT
			Min	Typ <sup>1</sup>	Max	Min	Typ <sup>1</sup>	Max	
T <sub>AA</sub> T <sub>CE</sub>	Access time Address Chip enable Output Output	Address Chip enable		30 15	40 25		30 15	70 40	ns
T <sub>CD</sub> T <sub>WD</sub>	Disable time Valid time Output Output	Chip enable Write enable		15 30	25 40		15 30	40 55	ns ns
T <sub>WSA</sub> T <sub>WHA</sub> T <sub>WSD</sub> T <sub>WHD</sub> T <sub>WSC</sub> T <sub>WHC</sub>	Setup and hold time Setup time Hold time Write enable Setup time Hold time Write enable Setup time Hold time Write enable	Address Address Data in CE		0 0 25 0 0 0	-5 -5 15 -5 -5		10 10 40 10 10 10	-5 -5 15 -5 -5	ns
T <sub>WP</sub>	Pulse width Write enable <sup>9</sup>			25	15		40	15	ns

NOTES

- All typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> + 25 °C.
- All voltage values are with respect to network ground terminal.
- Test each input one at a time.
- Measured with a logic low stored and V<sub>IL</sub> applied to CE1, CE2 and CE3.
- Measured with a logic high stored. Output sink current is supplied through a resistor to V<sub>CC</sub>.
- Measured with V<sub>IH</sub> applied to CE1, CE2 and CE3.
- Duration of the short-circuit should not exceed 1 second.
- I<sub>CC</sub> is measured with the Write enable and memory enable inputs grounded, all other inputs at 4.5V, and the output open.
- Minimum required to guarantee a Write into the slowest bit.

**2304-BIT BIPOLAR RAM (256 × 9)**

**82S212A (T.S.)**

**DESCRIPTION**

The organization of the 82S212A allows byte wide storage of data, including parity. Where parity is not required, the ninth bit can be used as a tag for each word stored. The 82S212A is ideal for scratch-pad, push-down stacks, buffer memories, and other internal memory applications in which space and performance requirements dictate a wide data path in favor of word depth.

The 82S212A data inputs and outputs are common (common I/O) with separate output disable (OD) line that allows ease of read/write operations using a common bus.

The 82S212A is available in the commercial temperature range. For the commercial temperature range (0°C to 75°C) specify N82S212AF or N.

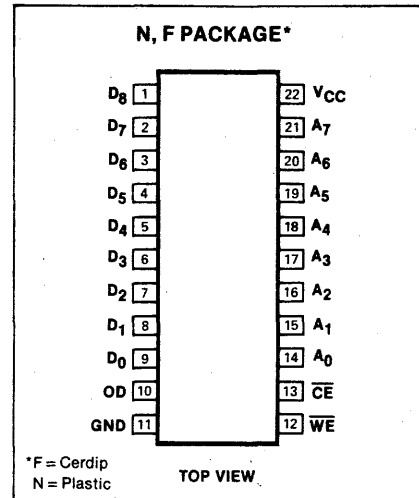
**FEATURES**

- Address access time:  
N82S212A: 35ns max
- Power dissipation: 0.3mW/bit
- Tri-state outputs
- Schottky clamped TTL

**APPLICATIONS**

- Cache memory
- Buffer storage
- Writable control store

**PIN CONFIGURATION**

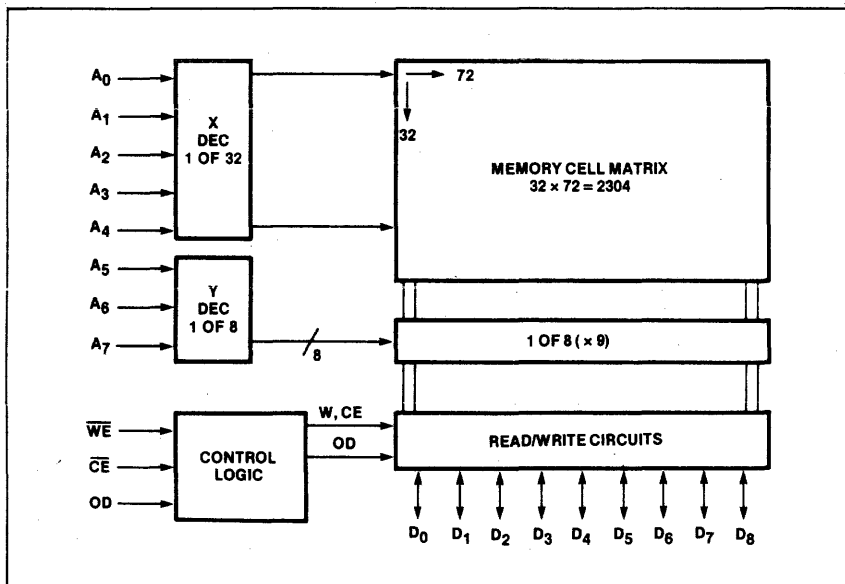


**TRUTH TABLE**

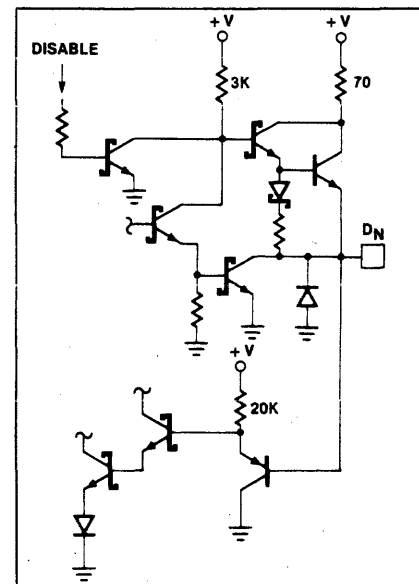
MODE	WE	CE	OD	DN IN/OUT
Disable output	X	X	1	High Z
Disable R/W	X	1	X	High Z
Write	0	0	1	Data in
Read	1	0	0	Data out

X = Don't care

**BLOCK DIAGRAM**



**TYPICAL I/O STRUCTURE**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT	
VCC	Supply voltage	+7	Vdc
VIN	Input voltage	+5.5	Vdc
VO	Off-state output voltage	+5.5	Vdc
TA	Operating	0 to +75	°C
TSTG	Storage	-65 to +150	

**2304-BIT BIPOLAR RAM (256 × 9)**

**82S212A (T.S.)**

**DC ELECTRICAL CHARACTERISTICS<sup>1</sup>** N82S212A: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V

PARAMETER	TEST CONDITIONS	N82S212A			UNIT	
		Min	Typ	Max		
V <sub>IL</sub> V <sub>IH</sub> V <sub>IC</sub>	Input voltage Low High Clamp <sup>2</sup>	V <sub>CC</sub> = Min V <sub>CC</sub> = Max V <sub>CC</sub> = Min, I <sub>IN</sub> = -12mA		2.0	0.8 -1.5	V
V <sub>OL</sub> V <sub>OH</sub>	Output voltage Low <sup>3</sup> High	V <sub>CC</sub> = Min, I <sub>OL</sub> = 8.0mA I <sub>OH</sub> = -2mA		2.4	0.5	V
I <sub>IL</sub> I <sub>IH</sub>	Input current Low High	V <sub>IN</sub> = 0.45V V <sub>IN</sub> = 5.5V			-100 25	μA
I <sub>OZ</sub> I <sub>OS</sub>	Output current Hi-Z state Short circuit <sup>4,5</sup>	$\overline{CE}$ = High or OD = High, V <sub>OUT</sub> = 5.5V $\overline{CE}$ = High or OD = High, V <sub>OUT</sub> = 0.5V $\overline{CE}$ = OD = Low, V <sub>OUT</sub> = 0V		-15	40 -100 -70	μA mA
I <sub>CC</sub>	V <sub>CC</sub> supply current	V <sub>CC</sub> = Max			135 185	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output	V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V V <sub>OUT</sub> = 2.0V			5 8	pF

**AC ELECTRICAL CHARACTERISTICS<sup>1</sup>** R<sub>1</sub> = 470Ω, R<sub>2</sub> = 1kΩ, C<sub>L</sub> = 30pF  
N82S212A: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V

PARAMETER	TO	FROM	N82S212A			UNIT	
			Min	Typ <sup>3</sup>	Max		
T <sub>AA</sub>	Access time Address	Output	Address			35	ns
T <sub>OE</sub> T <sub>CE</sub>	Enable time Output Output	Output Output	OD Chip enable			25 25	ns
T <sub>OD</sub> <sup>6</sup> T <sub>CD</sub> <sup>6</sup>	Disable time Output Output	Output Output	OD Chip enable			25 25	ns
T <sub>WP</sub>	Pulse width Write			25			ns
T <sub>WSC</sub> T <sub>WHC</sub>	Setup time Hold time	Write Chip enable	Chip enable Write	5 5			
T <sub>WSD</sub> T <sub>WHD</sub>	Setup time Hold Time	Write Data	Data Write	25 5			
T <sub>WSA</sub> T <sub>WHA</sub>	Setup time Hold time	Write Address	Address Write	5 5			
T <sub>SO</sub> T <sub>HO</sub>	Setup time (from disabled state) Hold time	Chip enable OD	OD Chip enable	50 50			

NOTES

- The operating ambient temperature ranges are guaranteed with transverse air flow exceeding 400 linear feet per minute and a 2 minute warmup.
- All voltages are with respect to network ground terminal.
- All typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.
- Measured on one pin at a time.
- Duration of I<sub>OS</sub> test should not exceed one second.
- Measured at a delta of 0.5V from Logic Level with R<sub>1</sub> = 750Ω, R<sub>2</sub> = 750Ω and C<sub>L</sub> = 5pF.

# 256-BIT BIPOLAR PROM (32 × 8)

# 82S23A (O.C.)/82S123A (T.S.)

## DESCRIPTION

The 82S23A and 82S123A are field programmable, which means that customer patterns are immediately available by following the fusing procedure given in this data manual. The 82S23A and 82S123A devices are supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

These devices include on-chip decoding and 1 chip enable input for memory expansion. They feature either open collector or tri-state outputs for optimization of word expansion in bused organizations.

Both 82S23A and 82S123A devices are available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S23A/123A, N or F, and for the military temperature range (-55°C to +125°C) specify S82S23A/123A, F or W.

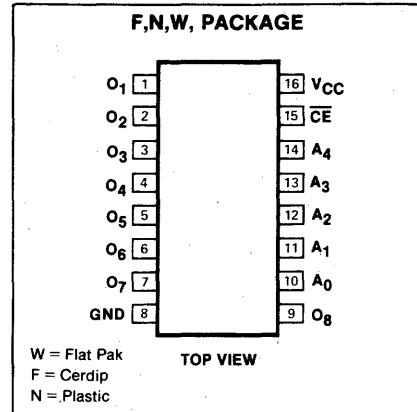
## FEATURES

- **Address access time:**  
N82S23A/123A: 25ns max  
S82S23A/123A: 35ns max
- **Power dissipation:** 1.3mW/bit typ
- **Input loading:**  
N82S23A/123A: -100µA max  
S82S23A/123A: -150µA max
- **On-chip address decoding**
- **Output options:**  
82S23A: Open collector  
82S123A: Tri-state
- **No separate fusing pins**
- **Unprogrammed outputs are low level**
- **Fully TTL compatible**

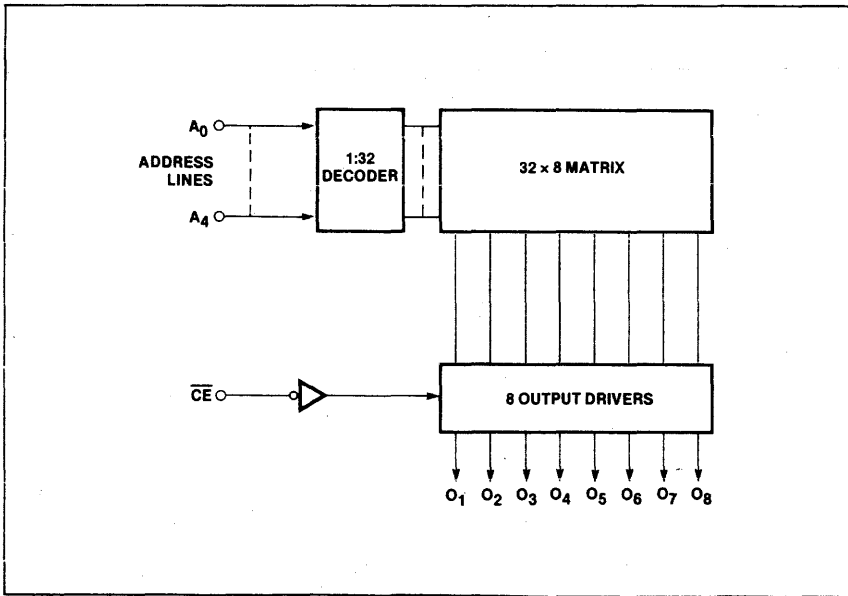
## APPLICATIONS

- Prototyping/volume production
- Sequential controllers
- Format conversion
- Hardwired algorithms
- Random logic
- Code conversion

## PIN CONFIGURATION



## LOGIC DIAGRAM



## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
V <sub>CC</sub>	Supply voltage	+7 Vdc
V <sub>IN</sub>	Input voltage	+5.5 Vdc
V <sub>OH</sub>	Output voltage	Vdc
V <sub>O</sub>	High (82S23) Off-state (82S123)	+5.5 +5.5
T <sub>A</sub>	Temperature range	°C
T <sub>STG</sub>	Operating	0 to +75
	Storage	-55 to +125 -65 to +150

Signetics

MEMORY



**256-BIT BIPOLAR PROM (32 × 8)**

**82S23A (O.C.)/82S123A (T.S.)**

**DC ELECTRICAL CHARACTERISTICS**

N82S23A/123A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S23A/123A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1, 2</sup>	N82S23A/123A			S82S23A/123A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$V_{IL}$ $V_{IH}$ $V_{IC}$	Input voltage Low High Clamp  $I_{IN} = -12\text{mA}$	2.0		0.8 -1.2	2.0		0.8 -1.2	V
$V_{OL}$ $V_{OH}$	Output voltage Low High  $\overline{CE} = \text{Low}$ $I_{OUT} = 16\text{mA}$ $I_{OUT} = -2\text{mA}$	2.4		0.45	2.4		0.5	V
$I_{IL}$ $I_{IH}$	Input current Low High  $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 50			-150 50	$\mu\text{A}$
$I_{OLK}$ $I_{OZ}$	Output current Leakage (82S23A) Hi-Z state (82S123A)  $\overline{CE} = \text{High}, V_{OUT} = 5.5\text{V}$ $\overline{CE} = \text{High}, V_{OUT} = 5.5\text{V}$ $\overline{CE} = \text{High}, V_{OUT} = 0.5\text{V}$			40 40 -40			50 50 -50	$\mu\text{A}$
$I_{OS}$	Short circuit (82S123A) <sup>3</sup> $\overline{CE} = \text{Low}, V_{OUT} = 0\text{V}, \text{High stored}$	-15		-90	-20		-100	mA
$I_{CC}$	$V_{CC}$ supply current $V_{CC} = \text{Max}$			96			110	mA
$C_{IN}$ $C_{OUT}$	Capacitance Input Output  $\overline{CE} = \text{High}, V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8			5 8		pF

**AC ELECTRICAL CHARACTERISTICS**

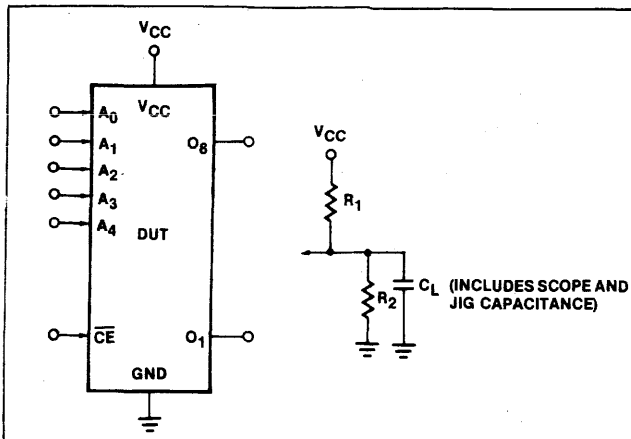
$R_1 = 270\Omega$ ,  $R_2 = 600\Omega$ ,  $C_L = 30\text{pF}$   
 N82S23A/123A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S23A/123A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S23A/123A			S82S23A/123A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$T_{AA}$ <sup>4</sup> $T_{CE}$	Access time Output Output	Address Chip enable		20	25 18			35 22	ns
$T_{CD}$	Disable time Output	Chip disable			18			22	ns

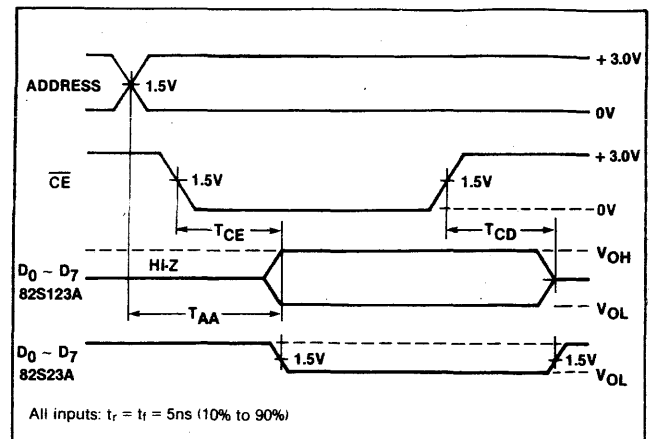
NOTES

- Positive current is defined as into the terminal referenced.
- All voltages with respect to network ground.
- Duration of short circuit should not exceed 1 second.
- Tested at an address cycle time of 1 $\mu\text{sec}$ .
- Typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

**TEST LOAD CIRCUIT**



**VOLTAGE WAVEFORM**



**1024-BIT BIPOLAR PROM (256 × 4)**

**82S126A (O.C.)/82S129A (T.S.)**

**DESCRIPTION**

The 82S126A and 82S129A are field programmable, which means that customer patterns are immediately available by following the fusing procedure given in this data manual. The 82S126A and 82S129A devices are supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

These devices include on-chip decoding and 2 chip enable input for ease of memory expansion. They feature either open collector or tri-state outputs for optimization of word expansion in bused organizations.

Both 82S126A and 82S129A devices are available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S126A/129A, F or N, and for the military temperature range (-55°C to +125°C) specify S82S126A/129A, F or R.

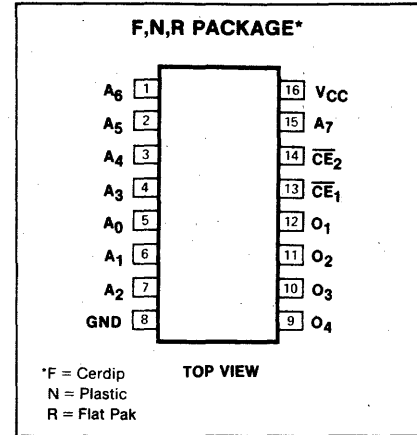
**FEATURES**

- **Address access time:**  
 N82S126A: 30ns max  
 N82S129A: 27ns max  
 S82S126A/129A: 35ns max
- **Power dissipation: 0.5mW/bit typ**
- **Input loading:**  
 N82S126A/129A: -100µA max  
 S82S126A/129A: -150µA max
- **On-chip address decoding**
- **Output options:**  
 82S126A: Open collector  
 82S129A: Tri-state
- **No separate fusing pins**
- **Unprogrammed outputs are low level**
- **Fully TTL compatible**

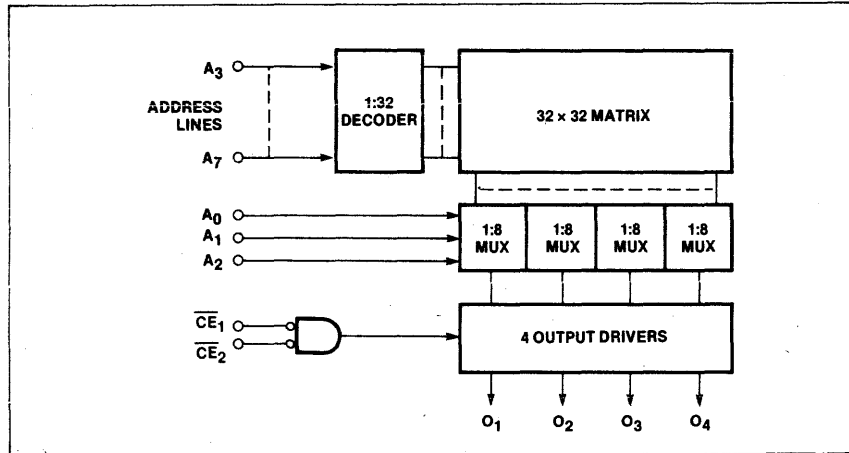
**APPLICATIONS**

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>OH</sub> Output voltage		Vdc
High (82S126)	+5.5	
Off-state (82S129)	+5.5	
T <sub>A</sub> Temperature range		°C
Operating		
N82S126A/129A	0 to +75	
S82S126A/129A	-55 to +125	
T <sub>STG</sub> Storage	-65 to +150	

Signetics

MEMORY

# 1024-BIT BIPOLAR PROM (256 x 4)

# 82S126A (O.C.)/82S129A (T.S.)

## DC ELECTRICAL CHARACTERISTICS

N82S126A/129A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S126A/129A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1, 2</sup>	N82S126A/129A			S82S126A/129A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$V_{IL}$ $V_{IH}$ $V_{IC}$	Input voltage Low High Clamp			0.8			0.8	V
$V_{OL}$ $V_{OH}$	Output voltage Low High (82S129)			0.45			0.5	V
$I_{IL}$ $I_{IH}$	Input current Low High			-100 40			-150 50	$\mu\text{A}$
$I_{OLK}$ $I_{OZ}$ $I_{OS}$	Output current Leakage (82S126A) Hi-Z state (82S129A) Short circuit (82S129A) <sup>3</sup>			40 40 -40			60 60 -60	$\mu\text{A}$ $\mu\text{A}$ mA
$I_{CC}$	$V_{CC}$ supply current			120			125	mA
$C_{IN}$ $C_{OUT}$	Capacitance Input Output			5 8			5 8	pF

## AC ELECTRICAL CHARACTERISTICS

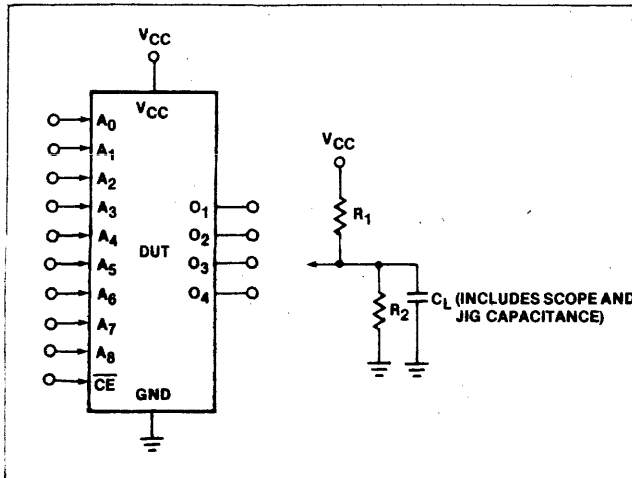
$R_1 = 270\Omega$ ,  $R_2 = 600\Omega$ ,  $C_L = 30\text{pF}$   
 N82S126A/129A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S126A/129A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S129A			N82S126A			S82S126A 82S129A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$T_{AA}$	Address access <sup>4</sup>	Output	Address		17	27		17	30		35	ns
$T_{CE}$	Enable access	Output	Chip enable		10	20		10	20		20	ns
$T_{CD}$	Output disable	Output	Chip enable		6	15		6	15		15	ns

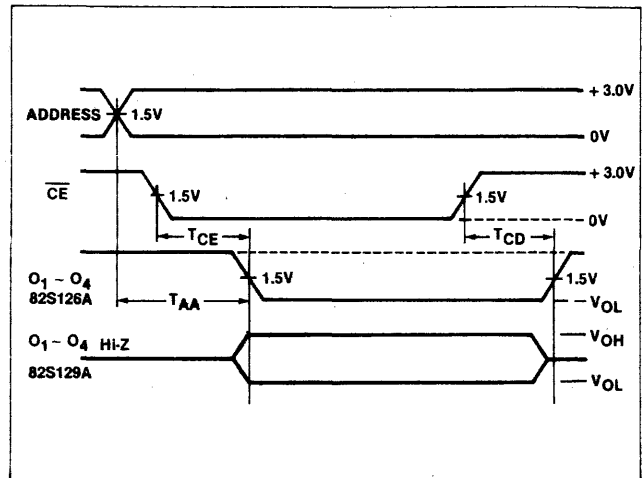
**NOTES**

- Positive current is defined as into the terminal referenced.
- All voltages with respect to network ground.
- Duration of short circuit should not exceed 1 second.
- Tested at an address cycle time of 1 $\mu\text{sec}$ .
- Typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

### TEST LOAD CIRCUIT



### VOLTAGE WAVEFORM



# 2048-BIT BIPOLAR PROM (512 x 4)

# 82S130A (O.C.)/82S131A (T.S.)

## DESCRIPTION

The 82S130A and 82S131A are field programmable, which means that customer patterns are immediately available by following the fusing procedure given in this data manual. The standard 82S130A and 82S131A are supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

These devices include on-chip decoding and 1 chip enable input for ease of memory expansion. They feature either open collector or tri-state outputs for optimization of word expansion in bused organizations.

Both 82S130A and 82S131A devices are available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S130A/131A, F or N, and for the military temperature range (-55°C to +125°C) specify S82S130A/131A, F or R.

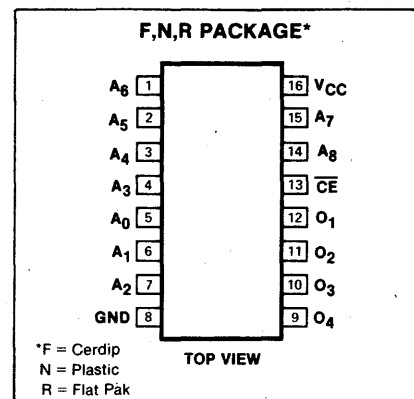
## FEATURES

- **Address access time:**  
N82S130A: 33ns max  
N82S131A: 30ns max  
S82S130A/131A: 35ns max
- **Power dissipation:** 0.3mW/bit typ
- **Input loading:**  
N82S130A/131A: -100µA max  
S82S130A/131A: -150µA max
- **On-chip address decoding**
- **Output options:**  
82S130A: Open collector  
82S131A: Tri-state
- **No separate fusing pins**
- **Unprogrammed outputs are low level**
- **Fully TTL compatible**

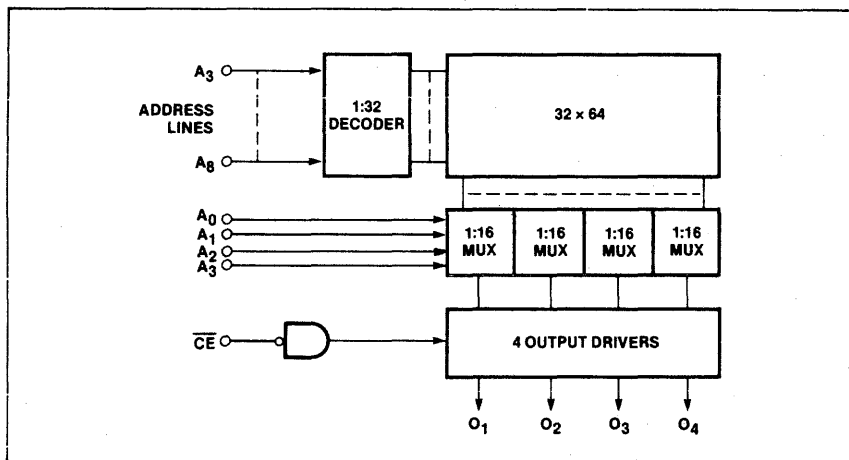
## APPLICATIONS

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

## PIN CONFIGURATION



## BLOCK DIAGRAM



## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
Output voltage		Vdc
V <sub>OH</sub> High (82S130)	+5.5	
V <sub>O</sub> Off-state (82S131)	+5.5	
Temperature range		°C
T <sub>A</sub> Operating	0 to +75	
N82S130A/131A	-55 to +125	
S82S130A/131A	-65 to +150	
T <sub>STG</sub> Storage		

**2048-BIT BIPOLAR PROM (512x4)**

**82S130A (O.C.)/82S131A (T.S.)**

**DC ELECTRICAL CHARACTERISTICS** N82S130A/131A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S130A/131A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

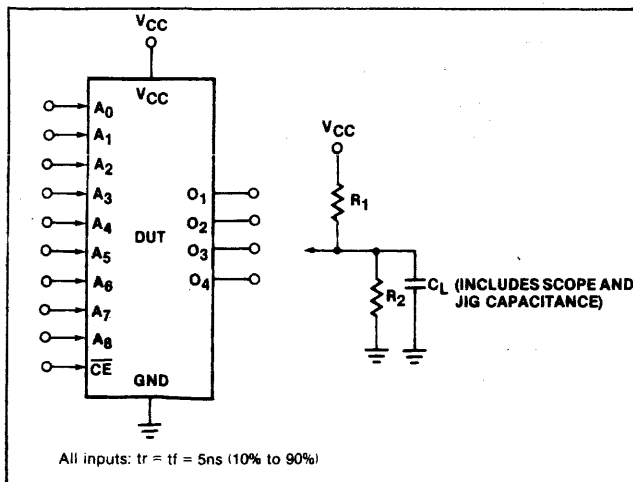
PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82S130A/131A			S82S130A/131A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
V <sub>IL</sub> V <sub>IH</sub> V <sub>IC</sub>	Input voltage Low High Clamp			0.8			0.8	V
V <sub>OL</sub> V <sub>OH</sub>	Output voltage Low High (82S131)			0.45			0.5	V
I <sub>IL</sub> I <sub>IH</sub>	Input current Low High			-100 40			-150 50	μA
I <sub>OLK</sub> I <sub>OZ</sub>	Output current Leakage (82S130A) Hi-Z state (82S131A)			40 40			60 60	μA
I <sub>OS</sub>	Short circuit (82S131A) <sup>3</sup>			-40			-60	mA
I <sub>CC</sub>	V <sub>CC</sub> supply current			140			140	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output			5 8			5 8	pF

**AC ELECTRICAL CHARACTERISTICS** R<sub>1</sub> = 270Ω, R<sub>2</sub> = 600Ω, C<sub>L</sub> = 30pF  
 N82S130A/131A:  $0^{\circ} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S130A/131A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

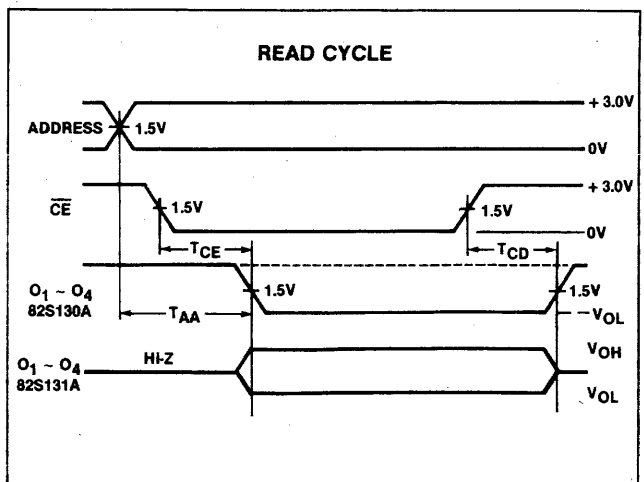
PARAMETER	TO	FROM	N82S131A			N82S130A			S82S130A N82S131A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
T <sub>AA</sub>	Address access <sup>4</sup>	Output	Address		18	30		18	33		35	ns
T <sub>CE</sub>	Enable access	Output	Chip enable		10	20		10	20		20	ns
T <sub>CD</sub>	Output disable	Output	Chip enable		6	15		6	15		15	ns

- NOTES
1. Positive current is defined as into the terminal referenced.
  2. All voltages with respect to network ground.
  3. Duration of short circuit should not exceed 1 second.
  4. Tested at an address cycle time of 1μsec.
  5. Typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.

**TEST LOAD CIRCUIT**



**VOLTAGE WAVEFORM**



**2048-BIT BIPOLAR PROM (256 × 8)**

**82LS135 (T.S.)**

**DESCRIPTION**

The 82LS135 is field-programmable, which means that custom patterns are immediately available by following the fusing procedure given in this data manual. The standard devices are supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

The 82LS135 includes on-chip decoding and two chip enable inputs for ease of memory expansion, and features tri-state outputs for optimization of word expansion in bused organizations.

The 82LS135 device is available in the commercial temperature range (0°C to 75°C), and is specified as N82LS135F or N.

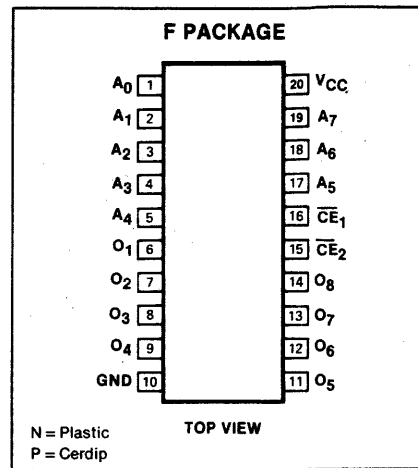
**FEATURES**

- Address access time: 100ns max.
- Power dissipation: 200µW/bit typ
- Input loading: -100µA max
- Two chip enable inputs
- On chip address decoding
- No separate fusing pins
- Fully TTL compatible

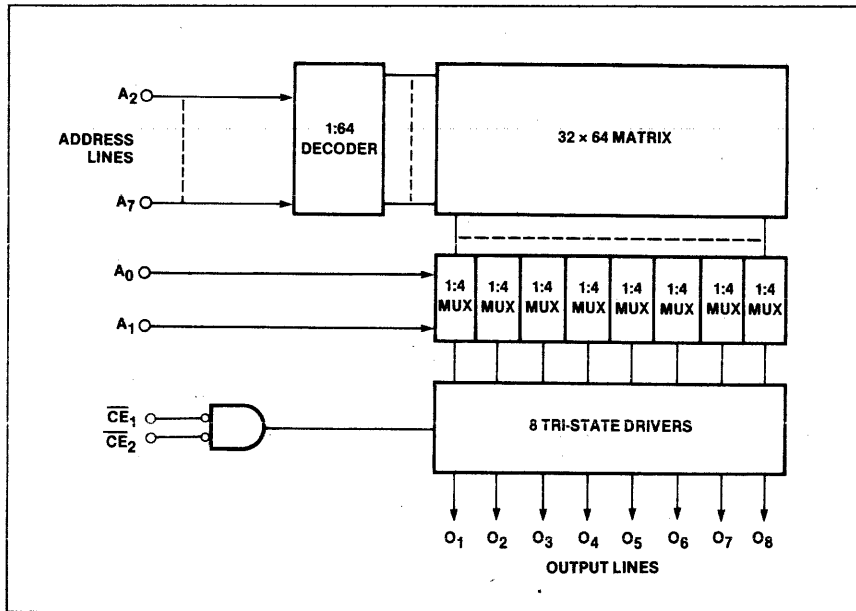
**APPLICATIONS**

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub>	+7	Vdc
V <sub>IN</sub>	+5.5	Vdc
V <sub>O</sub>	+5.5	Vdc
T <sub>A</sub>	-40 to +85	°C
T <sub>STG</sub>	-65 to +150	

Signetics  
MEMORY

# 2048-BIT BIPOLAR PROM (256 × 8)

# 82LS135 (T.S.)

## DC ELECTRICAL CHARACTERISTICS $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}, 4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	LIMITS			UNIT
		Min	Typ <sup>5</sup>	Max	
$V_{IL}$ $V_{IH}$ $V_{IC}$	Input voltage Low High Clamp $I_{IN} = -12\text{mA}$	2.0		.80 -1.2	V
$V_{OL}$ $V_{OH}$	Output voltage Low High $I_{OUT} = 16\text{mA}$ $\overline{CE}_1, \overline{CE}_2 = \text{Low}, I_{OUT} = -2\text{mA}, \text{High stored}$	2.4		.50	V
$I_{IL}$ $I_{IH}$	Input current Low High $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40	$\mu\text{A}$
$I_{OLK}$ $I_{OZ}$ $I_{OS}$	Output current Leakage Hi-Z state Short circuit <sup>3</sup> $\overline{CE}_1, \overline{CE}_2 = \text{High}, V_{OUT} = 5.5\text{V}$ $\overline{CE}_1, \overline{CE}_2 = \text{High}, V_{OUT} = 0.5\text{V}$ $\overline{CE}_1, \overline{CE}_2 = \text{High}, V_{OUT} = 5.5\text{V}$ $V_{OUT} = 0\text{V}$			40 -40 40 -75	$\mu\text{A}$ $\mu\text{A}$ mA
$I_{CC}$	$V_{CC}$ supply current $V_{CC} = \text{Max}$		80	100	mA
$C_{IN}$ $C_{OUT}$	Capacitance Input Output $V_{CC} = 5.0\text{V}$ $\overline{CE} = \text{High}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8		pF

## AC ELECTRICAL CHARACTERISTICS $R_1 = 10\text{ k}\Omega, R_2 = 20\text{ k}\Omega, C_L = 90\text{ pF}, 0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}, 4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$

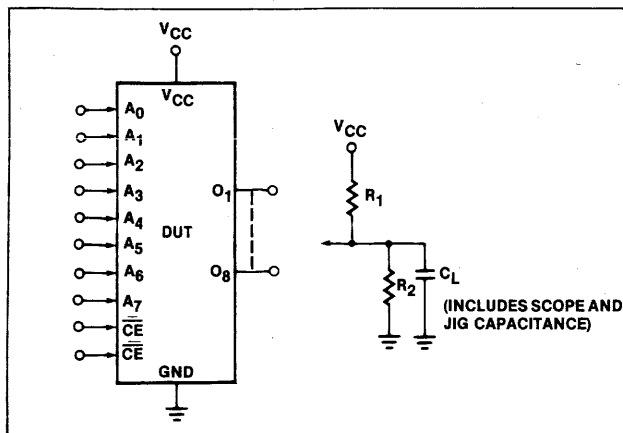
PARAMETER	TO	FROM	LIMITS			UNIT
			Min	Typ <sup>5</sup>	Max	
$T_{AA}$ <sup>4</sup> $T_{CE}$	Access time Output Output	Address Chip enable		70 30	100 50	ns
$T_{CD}$	Disable time Output	Chip disable		30	50	ns

**NOTES**

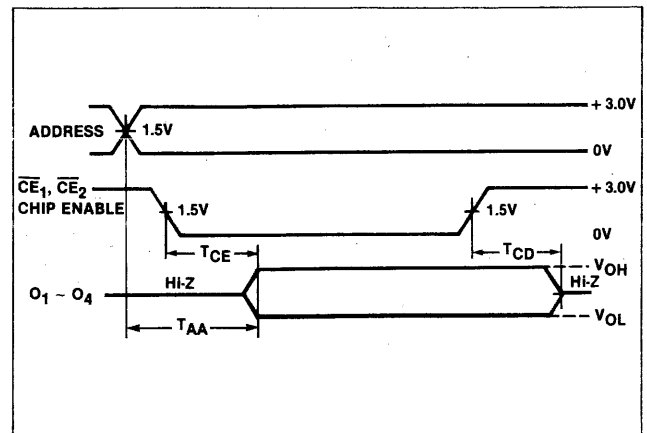
1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.

4. Tested at an address cycle time of  $1\mu\text{sec}$ .
5. Typical values are at  $V_{CC} = 5\text{V}, T_A = 25^{\circ}\text{C}$ .

### TEST LOAD CIRCUIT



### VOLTAGE WAVEFORM



# 4096-BIT BIPOLAR PROM (1024 × 4)

# 82S137A/82S137B (T.S.)

## DESCRIPTION

The 82S137 is field programmable, which means that custom patterns are immediately available by following the fusing procedure given in this data manual. The 82S137 is supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

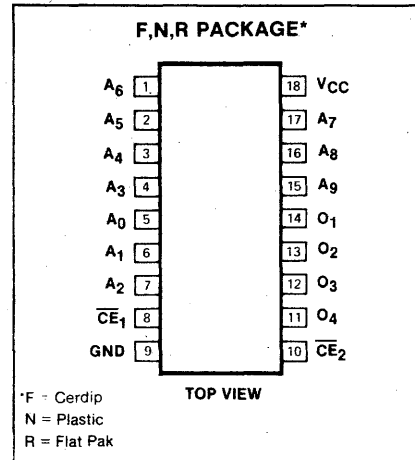
These devices include on-chip decoding and 2 chip enable inputs for ease of memory expansion. They feature tri-state outputs for optimization of word expansion in bused organizations.

The 82S137 device is available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S137AF or N, or N82S137BF or N, and for the military temperature range (-55°C to +125°C) specify S82S137AF or R.

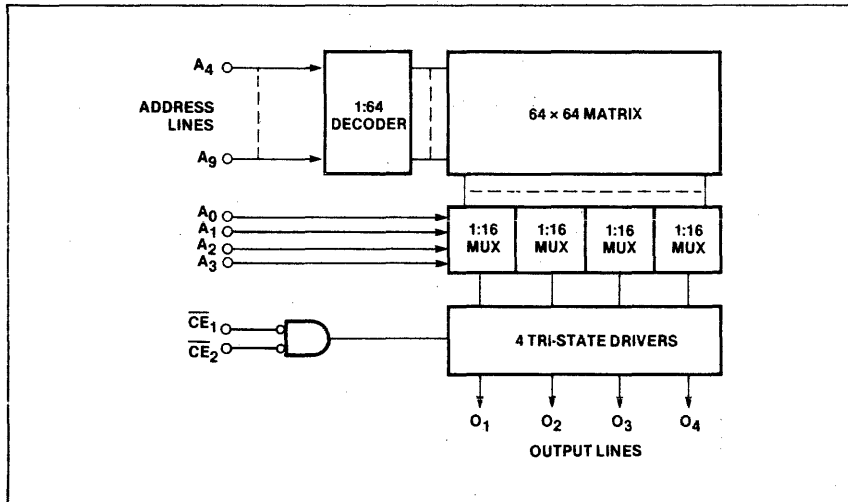
## FEATURES

- Address access time:  
 N82S137A: 45ns max  
 N82S137B: 35ns max  
 S82S137A: 70ns max
- Power dissipation: 13mW/bit typ
- Input loading:  
 N82S137: -100µA max  
 S82S137: -150µA max
- On-chip address decoding
- No separate fusing pins
- Unprogrammed outputs are low level
- Fully TTL compatible

## PIN CONFIGURATION



## BLOCK DIAGRAM



## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT	
V <sub>CC</sub>	Supply voltage	+7	Vdc
V <sub>IN</sub>	Input voltage	+5.5	Vdc
V <sub>O</sub>	Output voltage	+5.5	Vdc
	Off-state		
	Temperature range		°C
T <sub>A</sub>	Operating	0 to +75	
	N82S137	-55 to +125	
	S82S137	-65 to +150	
T <sub>STG</sub>	Storage		



# 4096-BIT BIPOLAR PROM (1024x4)

# 82S137A/82S137B (T.S.)

## DC ELECTRICAL CHARACTERISTICS

N82S137A/N82S137B:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S137A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82S137A/N82S137B			S82S137A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$V_{IL}$ $V_{IH}$ $V_{IC}$	Input voltage Low High Clamp $I_{IN} = -12\text{mA}$	2.0	-0.8	0.8 -1.2	2.0		0.8 -1.2	V
$V_{OL}$ $V_{OH}$	Output voltage Low High $\overline{CE}_{1,2} = \text{Low}$ $I_{OUT} = 16\text{mA}$ $I_{OUT} = -2\text{mA}$	2.4		0.45	2.4		0.5	V
$I_{IL}$ $I_{IH}$	Input current Low High $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40			-150 50	$\mu\text{A}$
$I_{OZ}$ $I_{OS}$	Output current Hi-Z state Short circuit <sup>3</sup> $\overline{CE}_{1,2} = \text{High}, V_{OUT} = 0.5\text{V}$ $\overline{CE}_{1,2} = \text{High}, V_{OUT} = 5.5\text{V}$ $\overline{CE}_{1,2} = \text{Low}, V_{OUT} = 0\text{V},$ Stored High	-15		40 -40 -70			60 -60 -85	$\mu\text{A}$ $\mu\text{A}$ mA
$I_{CC}$	$V_{CC}$ supply current $V_{CC} = \text{Max}$		85	140			150	mA
$C_{IN}$ $C_{OUT}$	Capacitance Input Output $\overline{CE}_{1,2} = \text{High}, V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8				5 8	pF

## AC ELECTRICAL CHARACTERISTICS

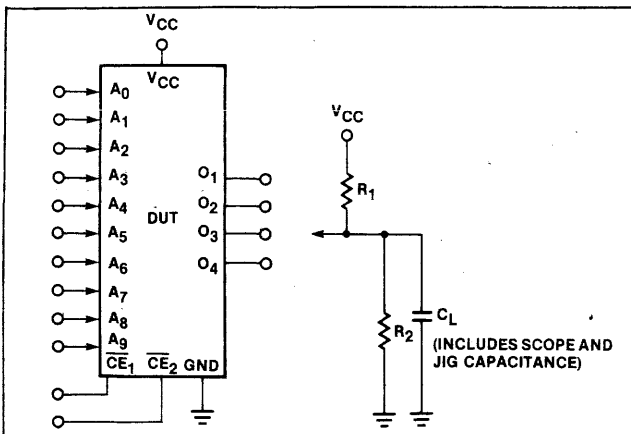
$R_1 = 270\Omega$ ,  $R_2 = 600\Omega$ ,  $C_L = 30\text{pF}$   
 N82S137A/N82S137B:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S137A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S137A			N82S137B			S82S137A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	Min	Typ	Max	
$T_{AA}$ <sup>4</sup> $T_{CE}$	Access time Output Output	Address Chip enable		35 15	45 30		30 25	35 25			70 40	ns
$T_{CD}$	Disable time Output	Chip disable		15	30			25			40	ns

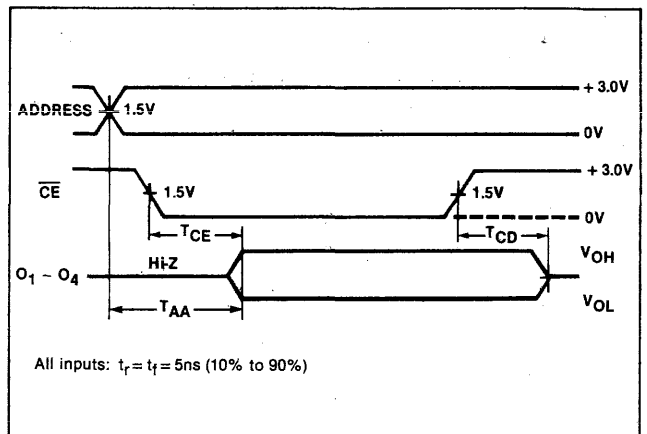
### NOTES

1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.
4. Tested at an address cycle time of 1 $\mu\text{sec}$ .
5. Typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

### TEST LOAD CIRCUIT



### VOLTAGE WAVEFORM



# 4096-BIT BIPOLAR PROM (512 × 8)

# 82S147A (T.S.)

### DESCRIPTION

The 82S147A is field-programmable, which means that custom patterns are immediately available by following the fusing procedure given in this data manual. The standard devices are supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

The 82S147A includes on-chip decoding and one chip enable input for ease of memory expansion, and features tri-state outputs for optimization of word expansion in bused organizations.

The 82S147A device is available in the commercial temperature range (0°C to +75°C), and is specified as N82S147A, F or N.

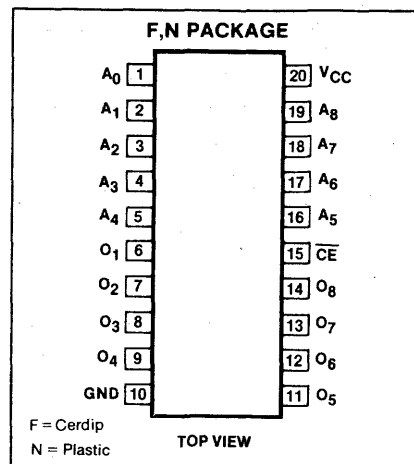
### FEATURES

- Address access time: 45ns max
- Power dissipation: 853mW max
- Input loading: -100µA max
- One chip enable input
- On chip address decoding
- No separate fusing pins
- Fully TTL compatible

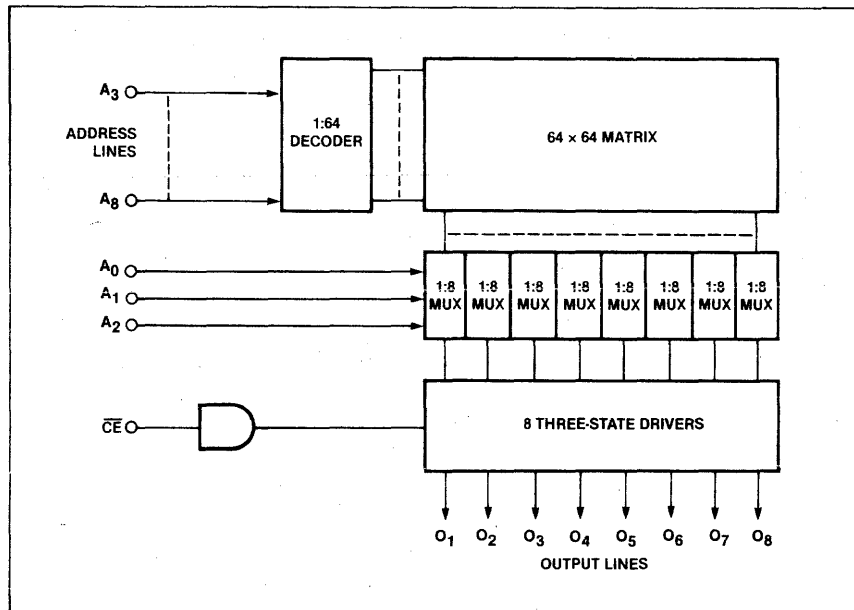
### APPLICATIONS

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

### PIN CONFIGURATION



### BLOCK DIAGRAM



### ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
V <sub>CC</sub> Power supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>O</sub> Output voltage	+5.5	Vdc
Off-state	+5.5	Vdc
Temperature range		°C
T <sub>A</sub> Operating	0 to +75	
T <sub>STG</sub> Storage	-65 to +150	

Signetics MEMORY

# 4096-BIT BIPOLAR PROM (512 × 8)

# 82S147A (T.S.)

## DC ELECTRICAL CHARACTERISTICS

N82S147A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S147:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITION <sup>1,2</sup>	N82S147A			S82S147A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$V_{IL}$ $V_{IH}$ $V_{IC}$	Input voltage Low High Clamp $I_{IN} = -12\text{mA}$			0.8			0.8	V
$V_{OL}$ $V_{OH}$	Output voltage Low High $\overline{CE} = \text{Low}$ $I_{OUT} = 9.6\text{mA}$ $I_{OUT} = -2\text{mA}$	2.4		0.45	2.4		0.5	V
$I_{IL}$ $I_{IH}$	Input current Low High $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40			-150 50	$\mu\text{A}$
$I_{OZ}$ $I_{OS}$	Output current Hi-Z state Short circuit <sup>3</sup> $\overline{CE} = \text{High}, V_{OUT} = 5.5\text{V}$ $\overline{CE} = \text{High}, V_{OUT} = 0.5\text{V}$ $\overline{CE} = \text{Low}, V_{OUT} = 0\text{V}$	-15		40 -40 -70	-15		-60 60 -85	$\mu\text{A}$ $\mu\text{A}$ mA
$I_{CC}$	$V_{CC}$ supply current $V_{CC} = \text{Max}$		125	155			165	mA
$C_{IN}$ $C_{OUT}$	Capacitance Input Output $\overline{CE} = \text{High}, V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8			5 8		pF

## AC ELECTRICAL CHARACTERISTICS

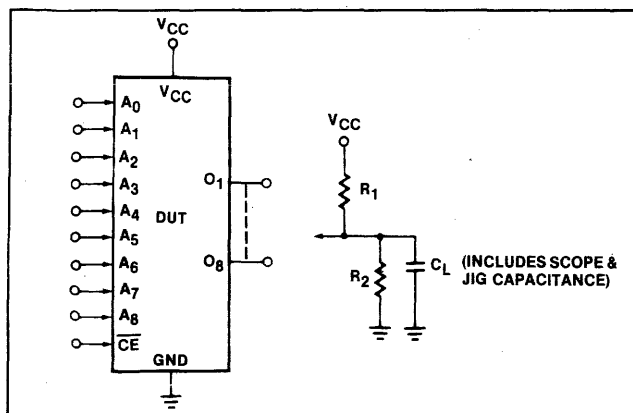
$R_1 = 470\Omega$ ,  $R_2 = 1\text{k}\Omega$ ,  $C_L = 30\text{pF}$   
 N82S147A:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S147A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S147A			S82S147A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$T_{AA}$ <sup>4</sup> $T_{CE}$	Access time Output Output	Address Chip enable		45 20	45 30			60 35	ns
$T_{CD}$	Disable time Output	Chip disable		20	30			35	ns

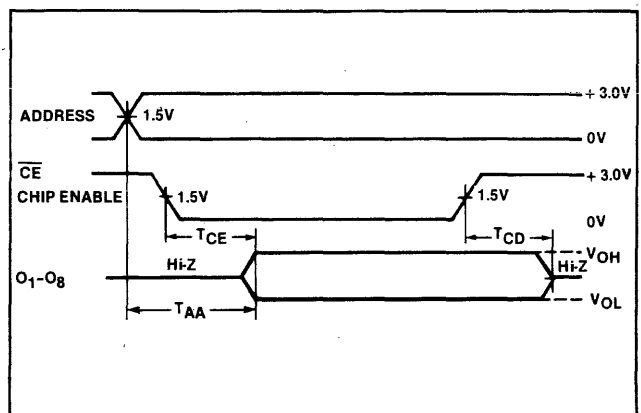
### NOTES

1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.
4. Tested at an address cycle time of  $1\mu\text{sec}$ .
5. Typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

### TEST LOAD CIRCUIT



### VOLTAGE WAVEFORM



**8192-BIT BIPOLAR PROM (1024 × 8)**

**82S181A/82S181B (T.S.)**

**DESCRIPTION**

The 82S181 is field-programmable, which means that custom patterns are immediately available by following the fusing procedure given in this data manual. The 82S181 is supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

This device includes on-chip decoding and 4 chip enable inputs for ease of memory expansion. It features 2 tri-state outputs for optimization of word expansion in bus-ed organizations.

The 82S181 is available in both the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S181AF or N or N82S181BF or N, and for the military temperature range (-55°C to +125°C) specify S82S181A, R, F, G.

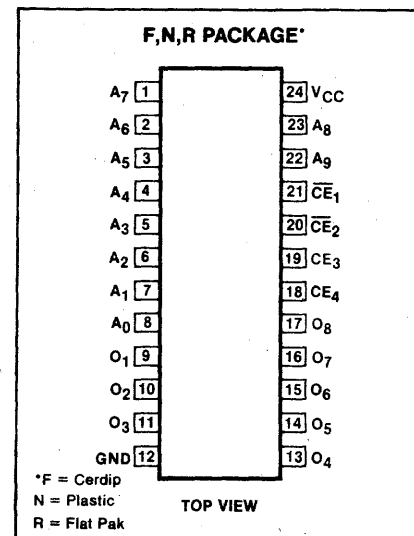
**FEATURES**

- Address access time:  
N82S181A: 55ns max  
N82S181B: 45ns max  
S82S181A: 80ns max
- Power dissipation: 76μW/bit typ
- Input loading:  
N82S181: -100μA max  
S82S181: -150μA max
- On-chip address decoding
- Output:  
82S181: tri-state
- No separate fusing pins
- Unprogrammed outputs are low level
- Fully TTL compatible

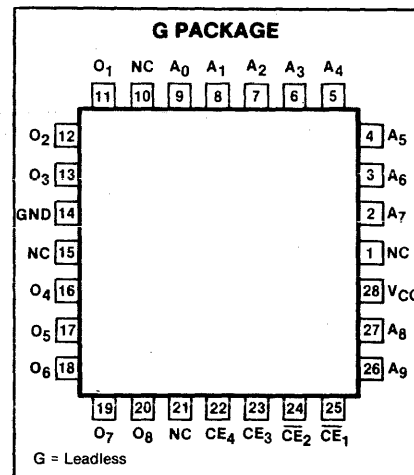
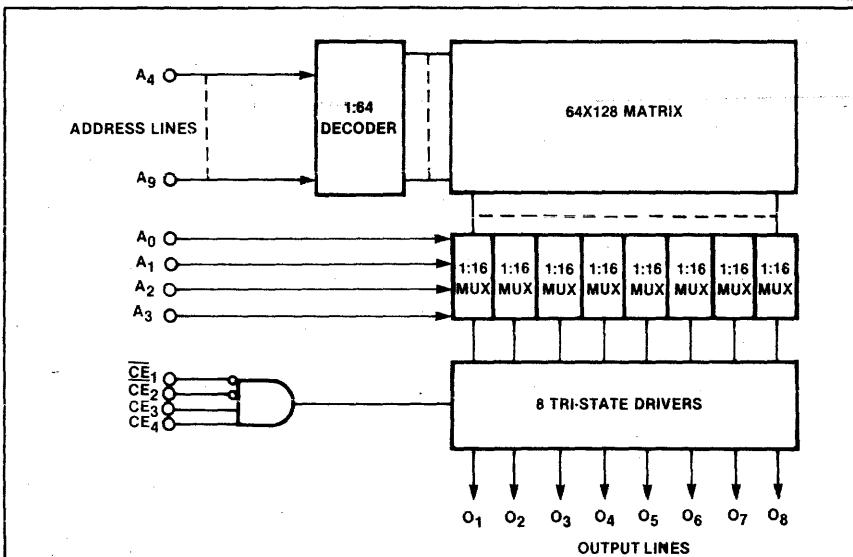
**APPLICATIONS**

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>O</sub> Output voltage		Vdc
V <sub>O</sub> Off-state	+5.5	
T <sub>A</sub> Temperature range		°C
Operating	N82S181	0 to +75
	S82S181	-55 to +125
T <sub>STG</sub> Storage	-65 to +150	

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**8192-BIT BIPOLAR PROM (1024x8)**

**82S181A/82S181B (T.S.)**

**DC ELECTRICAL CHARACTERISTICS** N82S181A/N82S181B:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S181A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82S181A/N82S181B			S82S181A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
V <sub>IL</sub> V <sub>IH</sub> V <sub>IC</sub>	Input voltage Low High Clamp  $I_{IN} = -12\text{mA}$			0.8			0.8	V
V <sub>OL</sub> V <sub>OH</sub>	Output voltage Low High  $\overline{CE}_{1,2} = \text{Low}, CE_{3,4} = \text{High}$ $I_{OUT} = 9.6\text{mA}$ $I_{OUT} = -2\text{mA}$	2.0	-0.8	-1.2	2.0		-1.2	V
I <sub>IL</sub> I <sub>IH</sub>	Input current Low High  $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40			-150 50	$\mu\text{A}$
I <sub>oz</sub> I <sub>os</sub>	Output current Hi-Z state Short circuit <sup>3</sup>  $\overline{CE}_{1,2} = \text{High}, CE_{3,4} = \text{Low}, V_{OUT} = 5.5\text{V}$ $\overline{CE}_{1,2} = \text{High}, CE_{3,4} = \text{Low}, V_{OUT} = 0.5\text{V}$ $\overline{CE}_{1,2} = \text{Low}, CE_{3,4} = \text{High}, V_{OUT} = 0\text{V}$ High Stored			40 -40 -70			60 -60 -85	$\mu\text{A}$ $\mu\text{A}$ mA
I <sub>CC</sub>	V <sub>CC</sub> supply current  $V_{CC} = \text{Max}$		125	175			185	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output  $\overline{CE}_{1,2} = \text{High}, V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8				5 8	pF

**AC ELECTRICAL CHARACTERISTICS**  $R_1 = 470\Omega$ ,  $R_2 = 1\text{k}\Omega$ ,  $C_L = 30\text{pF}$   
 N82S181A/N82S181B:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S181A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S181A			N82S181B			S82S181A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	Min	Typ	Max	
T <sub>AA</sub> <sup>4</sup> T <sub>CE</sub>	Access time	Output Output	Address Chip enable	45 20	55 35			45 30			80 45	ns
T <sub>CD</sub>	Disable time	Output	Chip disable	20	35			30			45	ns

**NOTES**

1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.
4. Tested at an address cycle time of 1 $\mu\text{sec}$ .
5. Typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

# 8192-BIT BIPOLAR PROM (2048X4)

# 82S185A/82S185B (T.S.)

## DESCRIPTION

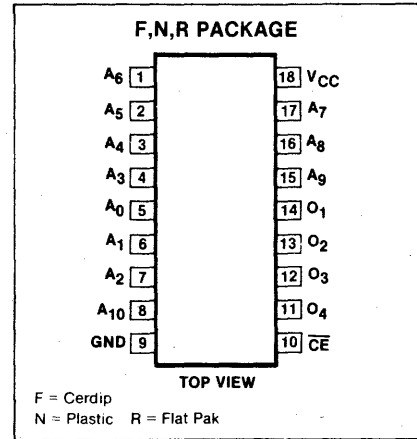
The 82S185 is field programmable, which means that custom patterns are immediately available by following the fusing procedure given in this data manual. The standard 82S185 is supplied with all outputs at logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

This device includes on-chip decoding and 1 chip enable input for memory expansion. It features tri-state outputs for optimization of word expansion in bused organizations.

The 82S185 device is available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S185A, F, N or N82S185B, F, or N, and for the military temperature range (-55°C to +125°C) specify S82S185A, F or R.

## FEATURES

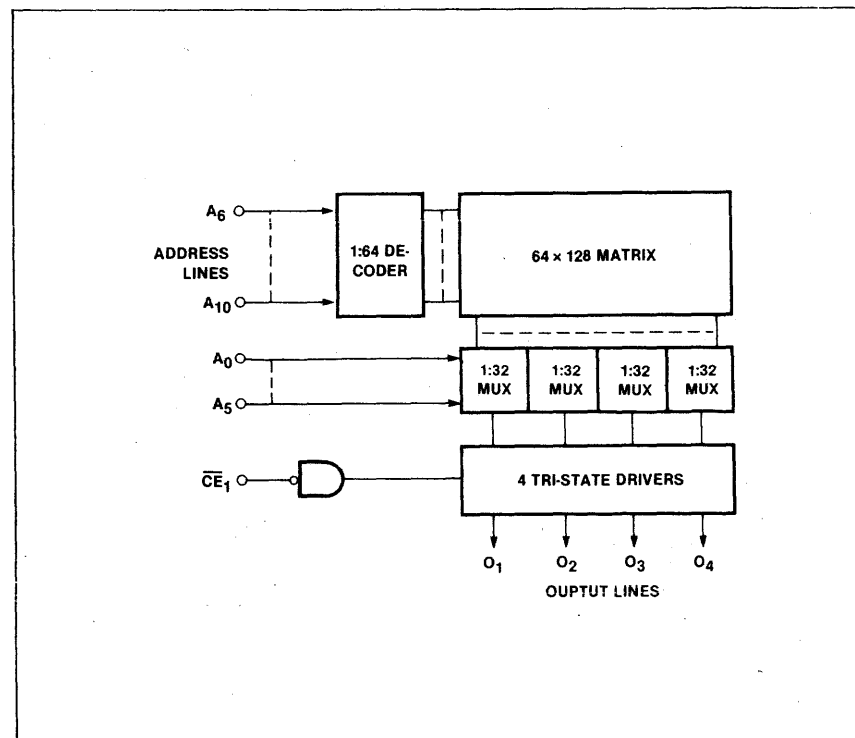
- **Low power dissipation:** 70 $\mu$ W/bit typ
- **Address access time:**  
 N82S185A: 50ns max  
 N82S185B: 45ns max  
 S82S185A: 80ns max
- **Input loading:**  
 N82S185: - 100 $\mu$ A max  
 S82S185: - 150 $\mu$ A max
- **On-chip address decoding**
- **No separate fusing pins**
- **Unprogrammed outputs are low level**
- **Fully TTL compatible**



## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT	
V <sub>CC</sub>	Supply voltage	+7	Vdc
V <sub>IN</sub>	Input voltage	+5.5	Vdc
V <sub>O</sub>	Output voltage	+5.5	Vdc
	Off-state		
	Temperature range		°C
T <sub>A</sub>	Operating	0 to +75	
	N82S185	-55 to +125	
	S82S185		
T <sub>STG</sub>	Storage	-65 to +150	

## BLOCK DIAGRAM



MEMORY Signetics

# 8192-BIT BIPOLAR PROM (2048X4)

# 82S185A/82S185B (T.S.)

**DC ELECTRICAL CHARACTERISTICS** N82S185A/N82S185B:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S185A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82S185A/B			S82S185A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
V <sub>IL</sub> V <sub>IH</sub> V <sub>IC</sub>	Input voltage Low High Clamp  $I_{IN} = -12\text{mA}$	2.0	-0.8	0.8 -1.2	2.0		0.8 -1.2	V
V <sub>OL</sub> V <sub>OH</sub>	Output voltage Low High  $\overline{CE} = \text{Low}$ $I_{OUT} = 16\text{mA}$ $I_{OUT} = -2\text{mA}$	2.4		0.45	2.4		0.5	V
I <sub>IL</sub> I <sub>IH</sub>	Input current Low High  $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40			-150 50	$\mu\text{A}$
I <sub>OZ</sub> I <sub>OS</sub>	Output current Hi-Z state Short circuit <sup>3</sup>  $\overline{CE} = \text{High}, V_{OUT} = 0.5\text{V}$ $\overline{CE} = \text{High}, V_{OUT} = 5.5\text{V}$ $\overline{CE} = \text{Low}, V_{OUT} = 0\text{V}$ High Stored			-40 40 -70		15	-60 60 -85	$\mu\text{A}$ mA
I <sub>CC</sub>	V <sub>CC</sub> supply current  $V_{CC} = \text{Max}$		110	155			160	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output  $\overline{CE} = \text{High}, V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8			5 8		pF

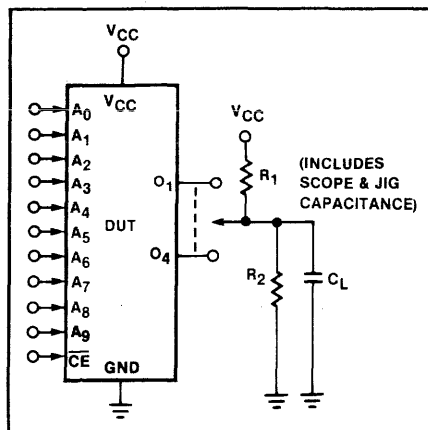
**AC ELECTRICAL CHARACTERISTICS**  $R_1 = 270\Omega$ ,  $R_2 = 600\Omega$ ,  $C_L = 30\text{pF}$   
 N82S185A/N82S185B:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S185A:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S185A			N82S185B			S82S185A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	Min	Typ	Max	
T <sub>AA</sub> <sup>4</sup> T <sub>CE</sub>	Access time Output Output	Address Chip enable		40 20	50 30			45 25			80 40	ns
TCD	Disable time Output	Chip disable		20	30			25			40	ns

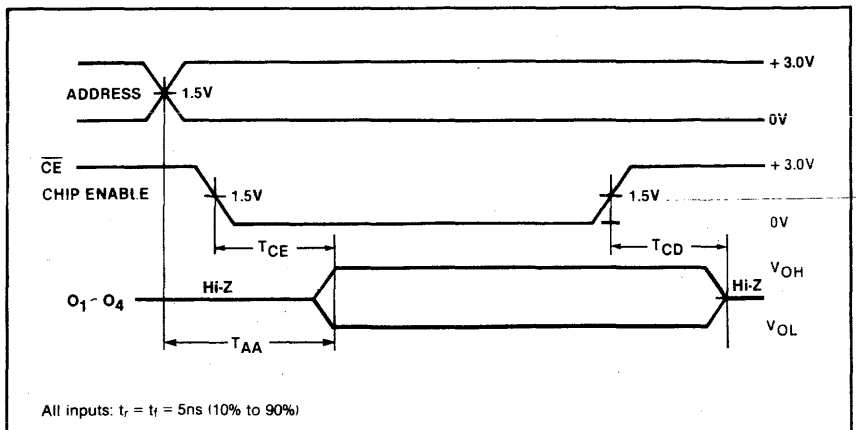
**NOTES**

1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.
4. Tested at an address cycle time of 1 $\mu\text{sec}$ .
5. Typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$

**TEST LOAD CIRCUIT**



**VOLTAGE WAVEFORM**



# 16,384-BIT BIPOLAR PROM (2048 × 8)

# 82S191A (T.S.)

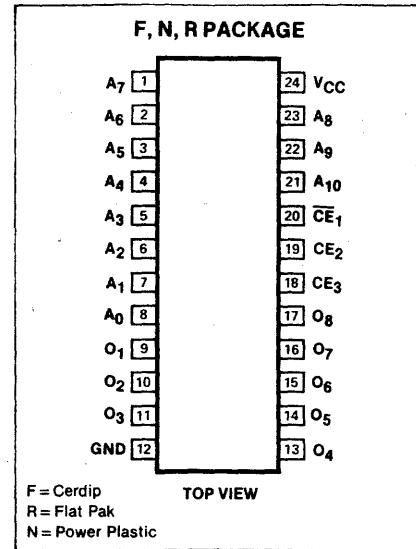
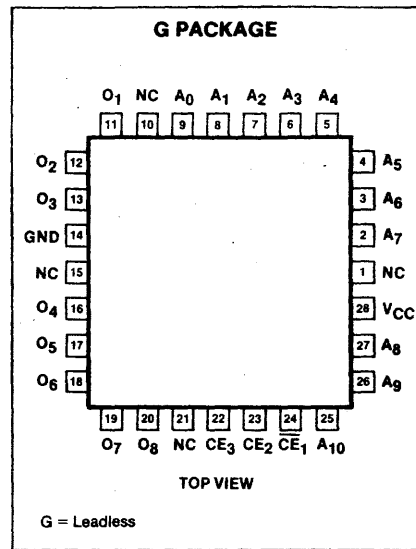
## DESCRIPTION

The 82S191 is field programmable, which means that custom patterns are immediately available by following the fusing procedure given in this data manual. The 82S191 is supplied with all outputs at a logical low. Outputs are programmed to a logic high level at any specified address by fusing a Ni-Cr link matrix.

This device includes on-chip decoding and 3 chip enable inputs for ease of memory expansion. It features tri-state outputs for optimization of word expansion in bused organizations.

The 82S191 device is available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S191A, F or N, and for the military temperature range (-55°C to +125°C) specify S82S191A, F, R or G.

## PIN CONFIGURATIONS



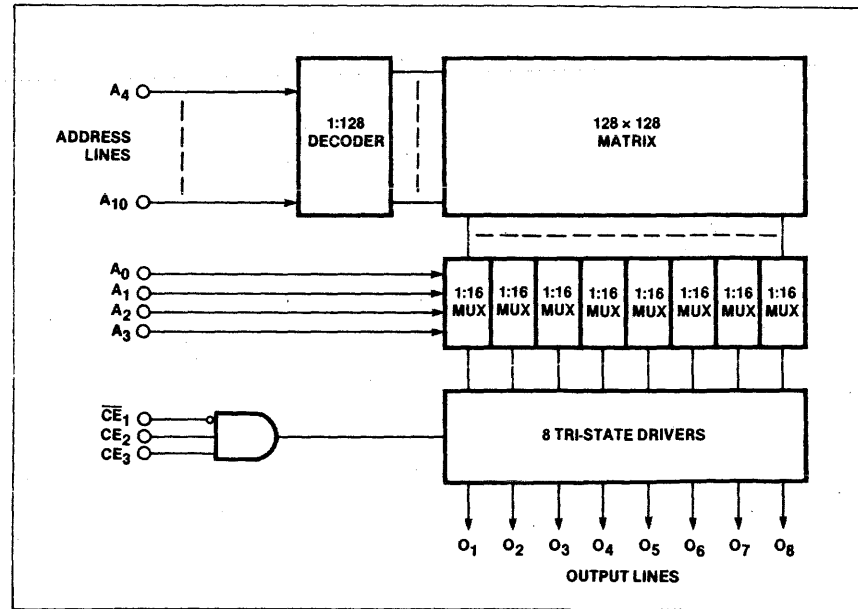
## FEATURES

- Address access time:
  - N82S191A: 55ns max
  - S82S181A: 70ns max
- Power dissipation: 40μW/bit typ
- Input loading:
  - N82S191A: -100μA max
  - S82S191A: -150μA max
- 3 chip enable inputs
- On-chip address decoding
- No separate fusing pins
- Unprogrammed outputs are low level
- Fully TTL compatible

## APPLICATIONS

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

## BLOCK DIAGRAM



## ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT	
V <sub>CC</sub>	Supply voltage	+7	Vdc
V <sub>IN</sub>	Input voltage	+5.5	Vdc
V <sub>O</sub>	Output voltage		Vdc
	Off-state	+5.5	
TA	Temperature range		°C
	Operating	0 to +75	
	N82S191		
	S82S191	-55 to +125	
T <sub>STG</sub>	Storage	-65 to +150	



**16,384-BIT BIPOLAR PROM (2048 × 8)**

**82S191A (T.S.)**

**DC ELECTRICAL CHARACTERISTICS**

N82S191A: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V

S82S191A: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82S191A			S82S191A			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ <sup>5</sup>	Max	
V <sub>IL</sub> V <sub>IH</sub> V <sub>IC</sub>	Input voltage Low High Clamp			0.8			0.8	V
		2.0	-0.8	-1.2	2.0		-1.2	
	I <sub>IN</sub> = -12mA							
V <sub>OL</sub> V <sub>OH</sub>	Output voltage Low High			0.45			0.5	V
		2.4			2.4			
	CE <sub>1</sub> = Low, CE <sub>2,3</sub> = High I <sub>OUT</sub> = 9.6mA I <sub>OUT</sub> = -2mA							
I <sub>IL</sub> I <sub>IH</sub>	Input current Low High			-100 40			-150 50	μA
	V <sub>IN</sub> = 0.45V V <sub>IN</sub> = 5.5V							
I <sub>OZ</sub>	Output current Hi-Z state			-40 40			-60 60	μA
I <sub>OS</sub>	Short circuit <sup>3</sup>			-70			-85	mA
	CE <sub>1</sub> = High, CE <sub>2,3</sub> = Low, V <sub>OUT</sub> = 0.5 CE <sub>1</sub> = High, CE <sub>2,3</sub> = Low, V <sub>OUT</sub> = 5.5 CE <sub>1</sub> = Low, CE <sub>2,3</sub> = High, V <sub>OUT</sub> = 0V	-15			-15			
I <sub>CC</sub>	V <sub>CC</sub> supply current		130	175			185	mA
	V <sub>CC</sub> = Max							
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output		5 8			5 8		pF
	V <sub>CC</sub> = 5.0V V <sub>IN</sub> = 2.0V V <sub>OUT</sub> = 2.0V							

**AC ELECTRICAL CHARACTERISTICS**

R<sub>1</sub> = 470Ω, R<sub>2</sub> = 1kΩ, C<sub>L</sub> = 30pF

N82S191A: 0°C ≤ T<sub>A</sub> ≤ +75°C, 4.75V ≤ V<sub>CC</sub> ≤ 5.25V

S82S191A: -55°C ≤ T<sub>A</sub> ≤ +125°C, 4.5V ≤ V<sub>CC</sub> ≤ 5.5V

PARAMETER	TO	FROM	N82S191A			S82S191A			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
T <sub>AA</sub> <sup>4</sup> T <sub>CE</sub>	Access time Output Output	Address Chip enable	40 20	55 30			70 35	ns	
T <sub>CD</sub>	Disable time Output	Chip disable	20	30			35	ns	

NOTES:

1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.

4. Tested at an address cycle time of 1μsec.
5. Typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.

**16,384-BIT BIPOLAR PROM (4096 × 4)**

**82HS195(T.S.)**

**Advance Information**

**DESCRIPTION**

The 82HS195 is field programmable, which means that custom patterns are immediately available by following the Generic II fusing procedure. The standard 82HS195 is supplied with all outputs at logical high. Outputs are programmed to a logic low level at any specified address by fusing a programmable matrix.

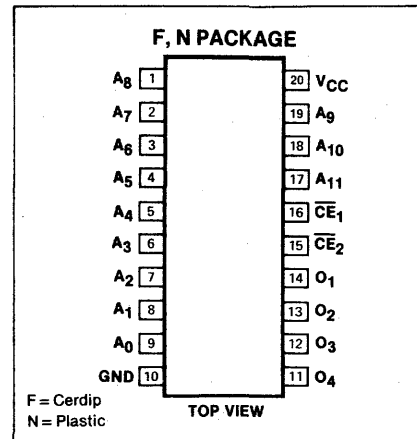
This device includes on-chip decoding and 2 chip enable inputs for memory expansion. It features tri-state outputs for optimization of word expansion in bused organizations.

The 82HS195 device is available in the commercial and military temperature ranges. For the commercial temperature range (0°C to +75°C) specify N82S195, F or N, and for the military temperature range (-55°C to +125°C) specify S82S195, F.

**FEATURES**

- Low power dissipation: 35µW/bit typ
- Address access time:  
N82HS195: 30ns max  
S82HS195: 40 ns max
- Input loading:  
N82HS195: -100µA max  
S82HS195: -150µA max
- On-chip address decoding
- No separate fusing pins
- Unprogrammed outputs are high level
- Fully TTL compatible

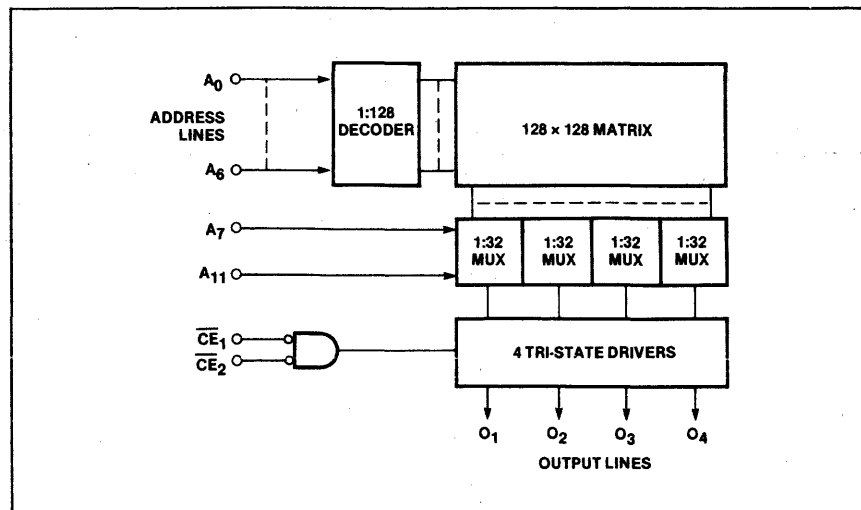
**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>O</sub> Output voltage		Vdc
V <sub>O</sub> Off-state	+5.5	
T <sub>A</sub> Temperature range		°C
Operating	N82HS195	0 to +75
	S82HS195	-55 to +125
T <sub>STG</sub> Storage	-65 to +150	

**BLOCK DIAGRAM**



Signetics  
MEMORY

# 16,384-BIT BIPOLAR PROM (4096 x 4)

# 82HS195(T.S.)

## Advance Information

**DC ELECTRICAL CHARACTERISTICS** N82HS195:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82HS195:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82HS195			S82HS195			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$V_{IL}$ $V_{IH}$ $V_{IC}$	Input voltage <sup>1</sup> Low High Clamp $I_{IN} = -12\text{mA}$	2.0	-0.8	0.8 -1.2	2.0		0.8 -1.2	V
$V_{OL}$ $V_{OH}$	Output voltage <sup>1</sup> Low High $I_{OUT} = 16\text{mA}$ $I_{OUT} = -2\text{mA}$	2.4		0.45	2.4		0.5	V
$I_{IL}$ $I_{IH}$	Input current Low High $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40			-150 50	$\mu\text{A}$
$I_{OZ}$ $I_{OS}$	Output current Hi-Z state Short circuit <sup>3</sup> $\overline{CE}_1$ & $\overline{CE}_2 = \text{High}$ , $V_{OUT} = 0.5\text{V}$ $\overline{CE}_1$ & $\overline{CE}_2 = \text{High}$ , $V_{OUT} = 5.5\text{V}$ $\overline{CE}_1$ & $\overline{CE}_2 = \text{Low}$ , $V_{OUT} = 0\text{V}$ High Stored			-40 40 -70			-60 60 -85	$\mu\text{A}$ mA
$I_{CC}$	$V_{CC}$ supply current $V_{CC} = \text{Max}$		110	155			165	mA
$C_{IN}$ $C_{OUT}$	Capacitance Input Output $\overline{CE}_1$ & $\overline{CE}_2 = \text{High}$ , $V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8			5 8		pF

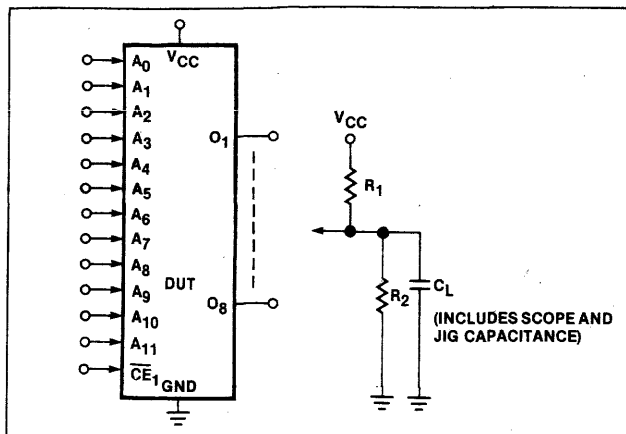
**AC ELECTRICAL CHARACTERISTICS**  $R_1 = 270\Omega$ ,  $R_2 = 600\Omega$ ,  $C_L = 30\text{pF}$   
 N82HS195:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82HS195:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82HS195			S82HS195			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
$T_{AA}$ <sup>4</sup> $T_{CE}$	Access time Output Output	Address Chip enable		25 20	30 20			40 25	ns
$T_{CD}$	Disable time Output	Chip disable		20	20			25	ns

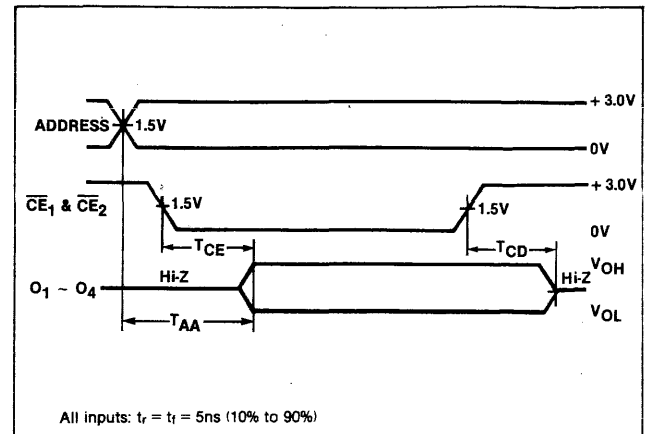
**NOTES**

- All voltage values are with respect to network ground terminal.
- Positive current is defined as into the terminal referenced.
- Duration of the short circuit should not exceed 1 second.
- Tested at an address cycle time of  $1\mu\text{sec}$ .
- All typical values are at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

### TEST LOAD CIRCUIT



### VOLTAGE WAVEFORM



**32,768-BIT BIPOLAR PROM (4096 × 8)**

**82HS321(T.S.)**

**Preliminary**

**DESCRIPTION**

The 82HS321 is field programmable, which means that custom patterns are immediately available by following the Generic II fusing procedure. The 82HS321 is supplied with all outputs at a logical high. Outputs are programmed to a logic low level at any specified address by fusing a programmable matrix.

This device includes on-chip decoding and 2 chip enable inputs for ease of memory expansion. It features tri-state outputs for optimization of word expansion in bused organizations.

The 82HS321 device is available in the commercial and military ranges. For the commercial temperature range (0°C to +75°C) specify N82HS321, F, N, and for the military temperature range (-55°C to +125°C) specify S82HS321, F, R or G.

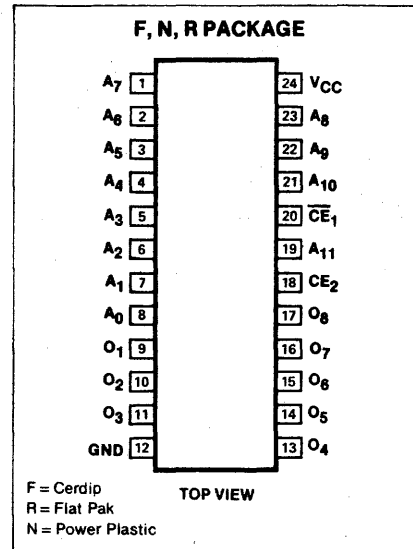
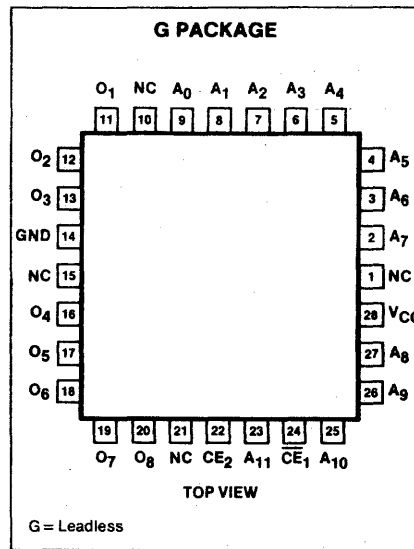
**FEATURES**

- Address access time:
  - N82HS321: 45ns max
  - S82HS321: 60ns max
- Power dissipation: 20μW/bit typ
- Input loading:
  - N82HS321: -100μA max
  - S82HS321: -150μA max
- 2 chip enable inputs
- On-chip address decoding
- No separate fusing pins
- Unprogrammed outputs are high level
- Fully TTL compatible

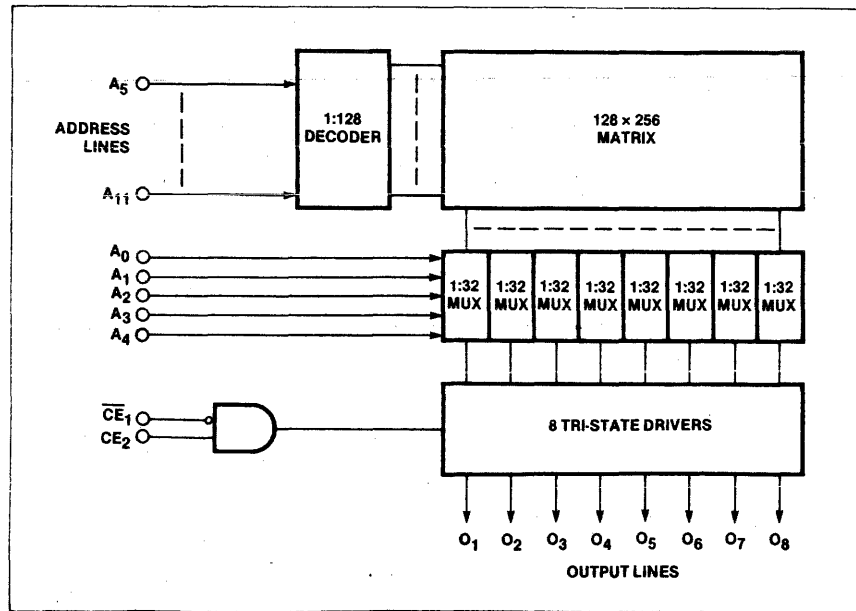
**APPLICATIONS**

- Prototyping/volume production
- Sequential controllers
- Microprogramming
- Hardwired algorithms
- Control store
- Random logic
- Code conversion

**PIN CONFIGURATIONS**



**BLOCK DIAGRAM**



**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING	UNIT
V <sub>CC</sub> Supply voltage	+7	Vdc
V <sub>IN</sub> Input voltage	+5.5	Vdc
V <sub>O</sub> Output voltage		Vdc
V <sub>O</sub> Off-state	+5.5	
T <sub>A</sub> Temperature range		°C
Operating	0 to +75	
N82HS321		
S82HS321	-55 to +125	
T <sub>STG</sub> Storage	-65 to +150	

Signetics

MEMORY

# 32,768-BIT BIPOLAR PROM (4096 x 8)

# 82HS321(T.S.)

**Preliminary**

**DC ELECTRICAL CHARACTERISTICS** N82S321:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S321:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TEST CONDITIONS <sup>1,2</sup>	N82S321			S82S321			UNIT
		Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
V <sub>IL</sub> V <sub>IH</sub> V <sub>IC</sub>	Input voltage Low High Clamp  $I_{IN} = -12\text{mA}$	2.0	-0.8	0.8 -1.2	2.0		0.8 -1.2	V
V <sub>OL</sub> V <sub>OH</sub>	Output voltage Low High  $\overline{CE}_1 = \text{Low}, CE_2 = \text{High}$ $I_{OUT} = 9.6\text{mA}$ $I_{OUT} = -2\text{mA}$	2.4		0.45	2.4		0.5	V
I <sub>IL</sub> I <sub>IH</sub>	Input current Low High  $V_{IN} = 0.45\text{V}$ $V_{IN} = 5.5\text{V}$			-100 40			-150 50	$\mu\text{A}$
I <sub>OZ</sub>	Output current Hi-Z state  $\overline{CE}_1 = \text{High}, CE_2 = \text{Low},$ $V_{OUT} = 0.5$ $\overline{CE}_1 = \text{High}, CE_2 = \text{Low},$ $V_{OUT} = 5.5$			-40 40			-60 60	$\mu\text{A}$
I <sub>OS</sub>	Short circuit <sup>3</sup>  $\overline{CE}_1 = \text{Low}, CE_2 = \text{High},$ $V_{OUT} = 0\text{V}$	-15		-70	-15		-85	mA
I <sub>CC</sub>	V <sub>CC</sub> supply current  $V_{CC} = \text{Max}$		130	175			185	mA
C <sub>IN</sub> C <sub>OUT</sub>	Capacitance Input Output  $\overline{CE}_1 = \text{High}, CE_2 = \text{Low},$ $V_{CC} = 5.0\text{V}$ $V_{IN} = 2.0\text{V}$ $V_{OUT} = 2.0\text{V}$		5 8			5 8		pF

**AC ELECTRICAL CHARACTERISTICS**  $R_1 = 470\Omega, R_2 = 1\text{k}\Omega, C_L = 30\text{pF}$   
 N82S321:  $0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$ ,  $4.75\text{V} \leq V_{CC} \leq 5.25\text{V}$   
 S82S321:  $-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$

PARAMETER	TO	FROM	N82S321			S82S321			UNIT
			Min	Typ <sup>5</sup>	Max	Min	Typ	Max	
T <sub>AA</sub> <sup>4</sup> T <sub>CE</sub>	Output	Address Chip enable			35 25			60 30	ns
T <sub>CD</sub>	Output	Chip disable			25			30	ns

**NOTES**

1. Positive current is defined as into the terminal referenced.
2. All voltages with respect to network ground.
3. Duration of short circuit should not exceed 1 second.
4. Tested at an address cycle time of 1 $\mu$ sec.
5. Typical values are at V<sub>CC</sub> = 5V, T<sub>A</sub> = 25°C.

# Signetics

## MOS MEMORY Products

### EPROMs

27C64

27C128

27128

### ROMs

2364-15/20

2664-15/20

26S64-30/45

23128-15/20

23128-25/30/45

23256A-20/25/30/45

**signetics**

a subsidiary of U.S. Philips Corporation

Signetics Corporation

811 E. Arques Avenue

P.O. Box 3409

Sunnyvale, California 94088-3409

Telephone 408/739-7700

**65,536-BIT CMOS EPROM (8192x8)**

**27C64**

**Advance Information**

**DESCRIPTION**

The Signetics 27C64 EPROM is a static complementary MOS (CMOS) 5V only, 65,536-bit electrically programmable and ultraviolet light erasable read only memory, organized as 8192 words by 8 bits.

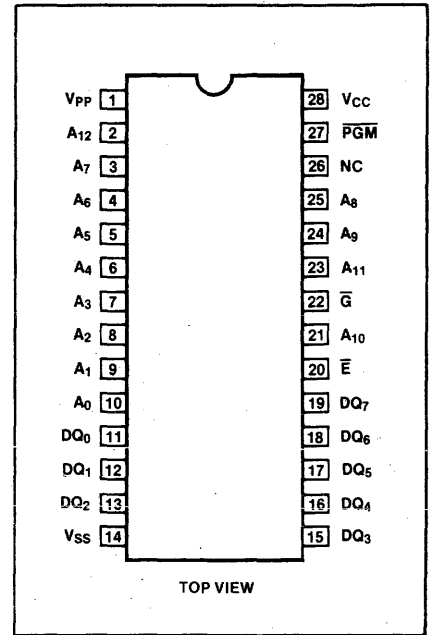
The 27C64 has a separate Output Enable control ( $\bar{G}$ ) from the Chip Enable Control ( $\bar{E}$ ), in order to eliminate bus contention in multiple bus systems.

The 27C64 typically consumes only 5 $\mu$ W power either in standby mode ( $\bar{E} = V_{CC}$ ) or in quiescent active mode ( $\bar{E} = V_{SS}$  pausing on one address location with  $V_{IN} = V_{CC}$  or  $V_{SS}$ ).

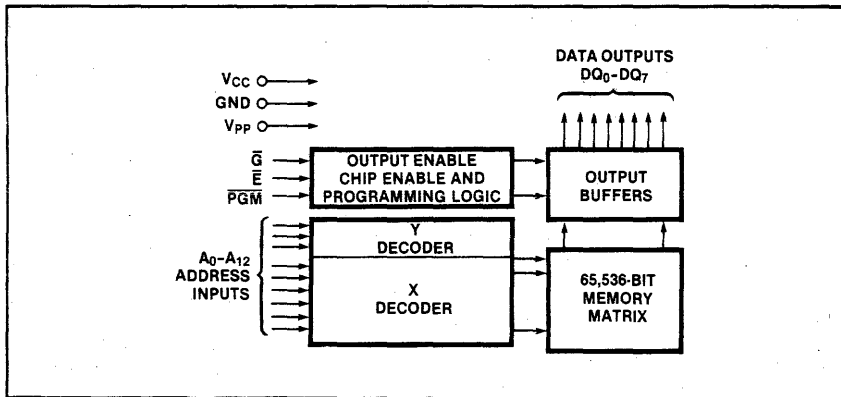
**FEATURES**

- Low CMOS power consumption — 5 $\mu$ W typ. (standby or quiescent active) 25mW typ. (active)
- 200ns access time
- TTL compatible
- Single +5V power supply
- Industry standard pinout
- Three-state outputs
- Two line controls
- Fast programming

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**PIN NAMES**

A <sub>0</sub> -A <sub>12</sub>	ADDRESSES
$\bar{E}$	Chip Enable
$\bar{G}$	Output Enable
DQ <sub>0</sub> -DQ <sub>7</sub>	Data Inputs/Outputs
PGM	Program Enable
N.C.	No Connect

**ABSOLUTE MAXIMUM RATINGS (Note 1)**

PARAMETER	RATING	UNIT
Temperature Range		
Storage	-65 to +125	°C
Bias	-10 to +80	
Operating Mode	0 to 70	
Program Mode	25 ± 5	
Applied Voltage with Respect to Ground Potential		
Input	-0.5 to +7	V
Output	-0.5 to +7	
Supply Voltage with Respect to Ground Potential		
	-0.5 to +6	V
	-0.5 to +22	

**NOTE**

1. Stresses beyond the limits listed under "Absolute Maximum Ratings" may cause damage to the device. These are stress ratings only. Functional operation at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rated conditions for extended periods of time may affect device reliability.

## 65,536-BIT CMOS EPROM (8192 × 8)

27C64

## Advance Information

DC ELECTRICAL CHARACTERISTICS DURING READ (Note 2)  $T_A = 0^\circ$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5.0\text{V} \pm 10\%$ , unless otherwise specified,  $V_{PP} = V_{CC}$

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			Min	Typ	Max	
$V_{IL}$	Input Voltage Low		-0.1		0.8	V
$V_{IH}$	Input Voltage High		2.0		$V_{CC} + 1$	V
$V_{OL}$	Output Voltage Low	$I_{OL} = 3.2\text{mA}$	—		0.45	V
$V_{OH}$	Output Voltage High	$I_{OH} = -400\mu\text{A}$	2.4		$V_{CC}$	V
$I_{LI}$	Input Load Current	$V_{IN} = 5.5\text{V}$ or $0.45\text{V}$	—		10	$\mu\text{A}$
$I_{LO}$	Output Leakage	$V_{OUT} = 5.5\text{V}$ or $0.45\text{V}$ , $\bar{E} = V_{CC}$	—		10	$\mu\text{A}$
$I_{CCS1}$	$V_{CC}$ Standby Current	$\bar{E} = V_{CC}$	—	1	100	$\mu\text{A}$
$I_{CCS2}$	$V_{CC}$ Standby Current	$\bar{E} = V_{IH}$	—		1	$\text{mA}$
$I_{CCQ1}$	$V_{CC}$ Active Current (Quiescent)	$\bar{E} = \bar{G} = V_{SS}$ All Addresses at either $V_{SS}$ or $V_{CC}$	—	1	100	$\mu\text{A}$
$I_{CCQ2}$	$V_{CC}$ Active Current (Quiescent)	$\bar{E} = \bar{G} = V_{IL}$ All Addresses at either $V_{IL}$ or $V_{IH}$		1.3	15	$\text{mA}$
$I_{CCA1}$	$V_{CC}$ Active Current (Operating)	$\bar{E} = \bar{G} = V_{SS}$ All addresses changing between $V_{SS}$ and $V_{CC}$	—	5	25	$\text{mA}$
$I_{CCA2}$	$V_{CC}$ Active Current (Operating)	$\bar{E} = \bar{G} = V_{IL}$ All addresses changing between $V_{IL}$ and $V_{IH}$		20	30	$\text{mA}$
$C_{IN}$	Input Capacitance (Note 3)	$V_{IN} = 0\text{V}$ , $f = 1\text{MHz}$		4	6	$\text{pF}$
$C_{OUT}$	Output Capacitance (Note 4)	$V_{OUT} = 0\text{V}$ , $f = 1\text{MHz}$		8	12	$\text{pF}$

## NOTES

- Typical values are for  $T_A = 25^\circ\text{C}$  and nominal power supply voltages.
- Typical values are for  $T_A = 25^\circ\text{C}$  and nominal power supplies.
- These parameters are periodically sampled and are not 100% tested.

AC ELECTRICAL CHARACTERISTICS DURING READ (Note 5)  $T_A = 0^\circ$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5\text{V} \pm 10\%$ , unless otherwise specified,  $V_{PP} = V_{CC}$

STANDARD SYMBOL	COMMON SYMBOL	PARAMETER	TEST CONDITIONS (Note 5)	LIMITS						UNIT
				27C64-20		27C64-25		27C64-30		
				Min	Max	Min	Max	Min	Max	
$t_{AVQV}$	$t_{ACC}$	Address Access Time	$\bar{E} = \bar{G} = V_{IL}$		200		250		300	ns
$t_{ELQV}$	$t_{CE}$	Chip Select Delay	$\bar{G} = V_{IL}$		200		250		300	ns
$t_{GLQV}$	$t_{OE}$	Output Enable Delay	$\bar{E} = V_{IL}$ (Note 6)	10	70	10	100	10	150	ns
$t_{EHQZ}$	$t_{DF}$	Chip Deselect Delay (Note 7)	$\bar{E} = V_{IL}$ $t_{DF}$ is measured at $V_{OL} = 0.8$ to tri-state (High-Z) transition	0	60	0	90	0	130	ns
$t_{AXQZ}$	$t_{OH}$	Previous Output Data Holding Time	From Addresses, $\bar{E}$ or $\bar{G}$ whichever occurred first $\bar{E} = \bar{G} = V_{IL}$	0		0		0		ns

## NOTES

- AC TEST CONDITIONS:  
Output load 2 TTL gates and  $C_L = 100\text{pF}$   
Input rise and fall times:  $\leq 20\text{ns}$   
Input pulse levels: 0.8V to 2.2V  
Timing measurement reference levels:  
Inputs 1.5V  
Outputs 0.6V and 2.2V
- $G$  may be delayed up to  $t_{AVQV} - t_{GLQV}$  after the falling edge of  $\bar{E}$  without impact on  $t_{AVQV}$ .
- $t_{EHQZ}$  is specified from  $\bar{G}$  or  $\bar{E}$ , whichever occurs first.



65,536-BIT CMOS EPROM (8192x8)

27C64

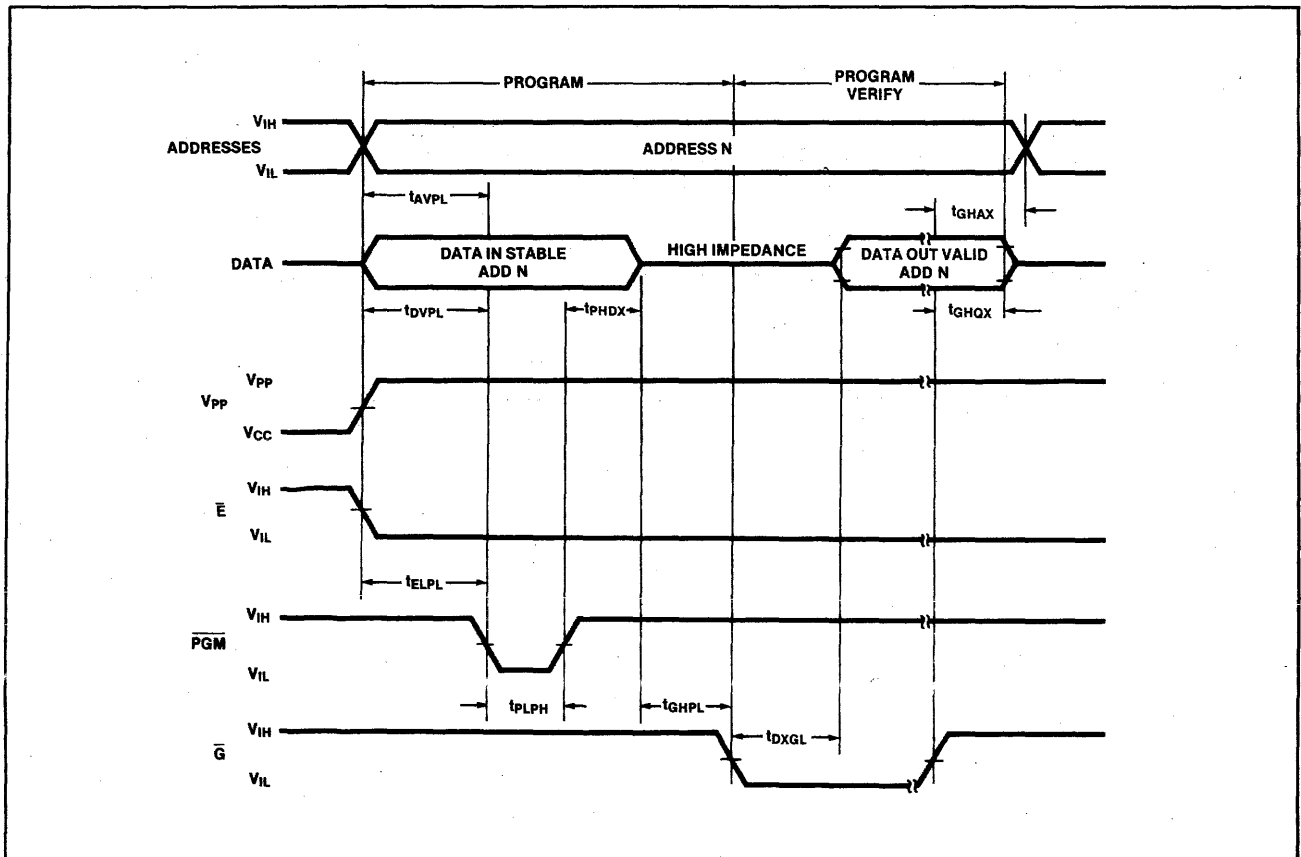
Advance Information

AC ELECTRICAL CHARACTERISTICS DURING PROGRAMMING (Note 10)  $T_A = 25 \pm 5^\circ\text{C}$ ,  $V_{CC} = 5V \pm 5\%$ ,  $V_{PP} = 21V \pm 0.5V$

STANDARD SYMBOL	COMMON SYMBOL	PARAMETER	TEST CONDITIONS(Note 2)	LIMITS			UNIT
				Min	Typ	Max	
$t_{AVPL}$	$t_{AS}$	Address Set-up Time		2			$\mu\text{S}$
$t_{DVPL}$	$t_{DS}$	Data Set-up Time		2			$\mu\text{S}$
$t_{PHDX}$	$t_{DH}$	Data Hold Time		2			$\mu\text{S}$
$t_{VHPL}$	$t_{VPPS}$	$V_{PP}$ Set-up Time		2			$\mu\text{S}$
$t_{ELPL}$	$t_{CES}$	$\overline{CE}$ Set-up Time		2			$\mu\text{S}$
$t_{PLPH}$	$t_{PW}$	$\overline{PGM}$ Pulse Width During Programming		25	50	55	mS
$t_{GHPL}$	$t_{OES}$	$\overline{OE}$ Set-up Time		2			$\mu\text{S}$
$t_{GHQZ}$	$t_{DF}$	Chip Deselect Delay (Note 9)	$\overline{CE} = V_{IL}$	0		130	ns
$t_{PHAX}$	$t_{AH}$	Address Hold Time					$\mu\text{S}$

Note:  
 9.  $t_{GHQZ}$  is specified from  $\overline{G}$  or  $\overline{E}$ , whichever occurs first.  
 10. The input reference level is 1V for  $V_{IL}$  and 2V for  $V_{IH}$ .

TIMING DIAGRAM DURING PROGRAMMING



**65,536-BIT CMOS EPROM (8192 x 8)**

**27C64**

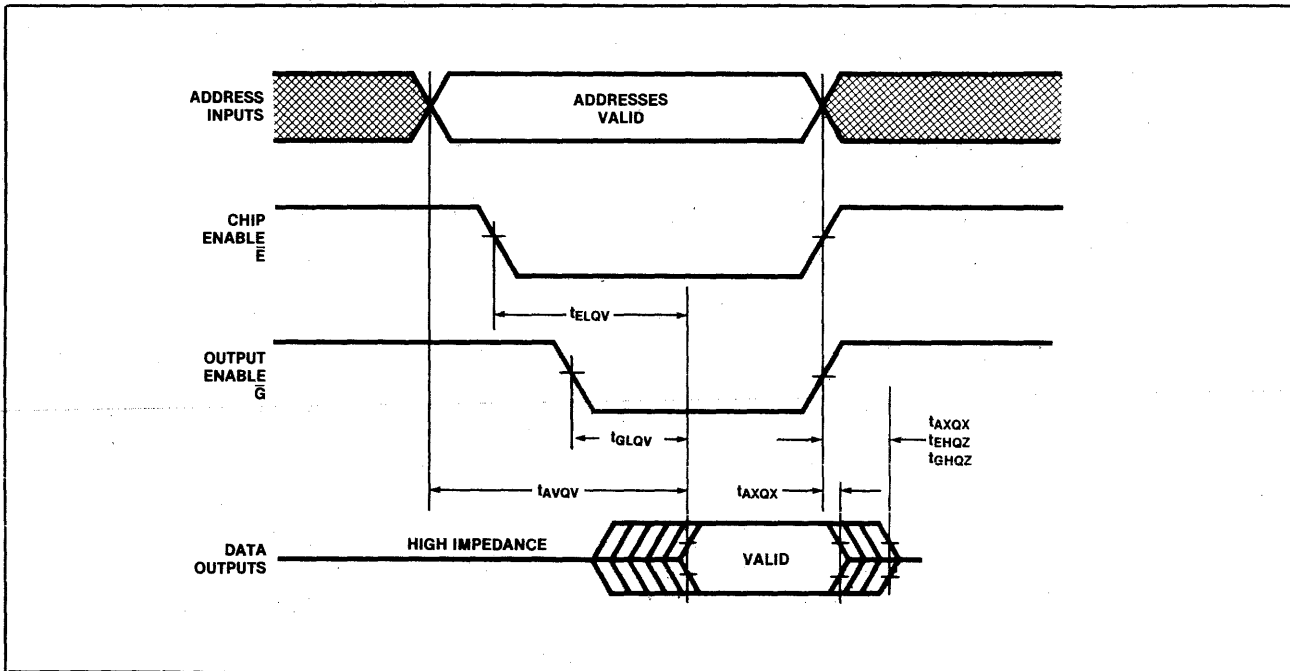
**Advance Information**

**MODE SELECTION**

MODE/PINS	$\bar{E}$ (20)	$\bar{G}$ (22)	PGM (27)	$V_{PP}$ (1)	$V_{CC}$ (28)	DQ <sub>0</sub> -DQ <sub>7</sub> (11-13, 15-19)
Read	$V_{IL}$	$V_{IL}$	$V_{IH}$	$V_{CC}$	$V_{CC}$	DQ <sub>OUT</sub>
Standby	$V_{IH}$	X	X	$V_{CC}$	$V_{CC}$	High Z
Program	$V_{IL}$	$V_{IH}$	$V_{IL}$	$V_{PP}$	$V_{CC}$	DQ <sub>IN</sub>
Program Verify	$V_{IL}$	$V_{IL}$	$V_{IH}$	$V_{PP}$	$V_{CC}$	DQ <sub>OUT</sub>
Program Inhibit	$V_{IH}$	X	X	$V_{PP}$	$V_{CC}$	High Z

X can be either  $V_{OL}$  or  $V_{OH}$ .

**TIMING DIAGRAM DURING READ**



**DC ELECTRICAL CHARACTERISTICS DURING PROGRAMMING** (Note 8)  $T_A = 25 \pm 5^\circ C$ ,  $V_{CC} = 5V \pm 5\%$ ,  $V_{PP} = 21V \pm 0.5V$

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			Min	Typ	Max	
$V_{IL}$	Input Voltage — Low		-0.1		0.8	V
$V_{IH}$	Input Voltage — High		2.0		$V_{CC} + 1$	V
$V_{OL}$	Output Voltage During Verify — Low	$I_{OL} = 3.2mA$			0.45	V
$V_{OH}$	Output Voltage During Verify — High	$I_{OH} = 400\mu A$	2.4			V
$I_{LI}$	Input Load Current	$V_{IN} = 5.25V$ or $V_{IL}$			10	$\mu A$
$I_{CCA3}$	$V_{CC}$ Supply Current — Active			20	30	mA
$I_{PP2}$	$V_{PP}$ Supply Current During Programming Pulse	$\bar{E} = \bar{PGM} = V_{IL}$ $V_{PP} = 21.5V, DQ_0-DQ_7 = V_{IL}$			30	mA

Note:  
8.  $V_{CC}$  must be applied before and removed after  $V_{PP}$ .  $V_{PP}$  should never exceed 21.5V (including transients). If  $\bar{PGM} = V_{IL}$ ,  $V_{PP}$  should not be switched.

MEMORY Signetics

**131,072-BIT CMOS EPROM (16,384 × 8)**

**27C128**

**Preview**

**DESCRIPTION**

The Signetics 27C128 EPROM is a static complementary MOS (CMOS) 5V only, 131,072-bit electrically programmable and ultraviolet light erasable read only memory, organized as words by bits.

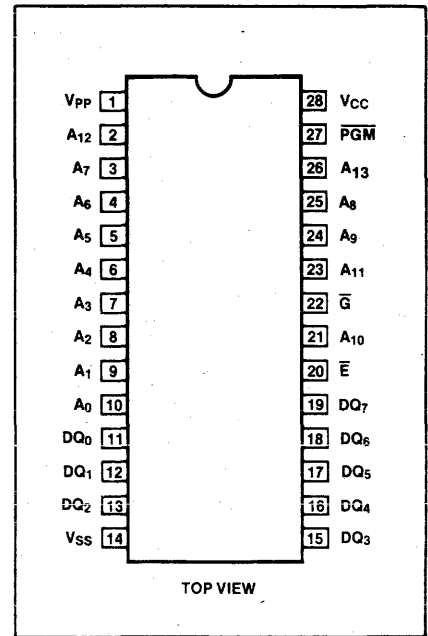
The 27C128 has a separate Output Enable control ( $\bar{G}$ ) from the Chip Enable Control ( $\bar{E}$ ), in order to eliminate bus contention in multiple bus systems.

The 27C128 typically consumes only 5 $\mu$ W power either in standby mode ( $\bar{E} = V_{CC}$ ) or in quiescent active mode ( $\bar{E} = V_{SS}$  pausing on one address location with  $V_{IN} = V_{CC}$  or  $V_{SS}$ ).

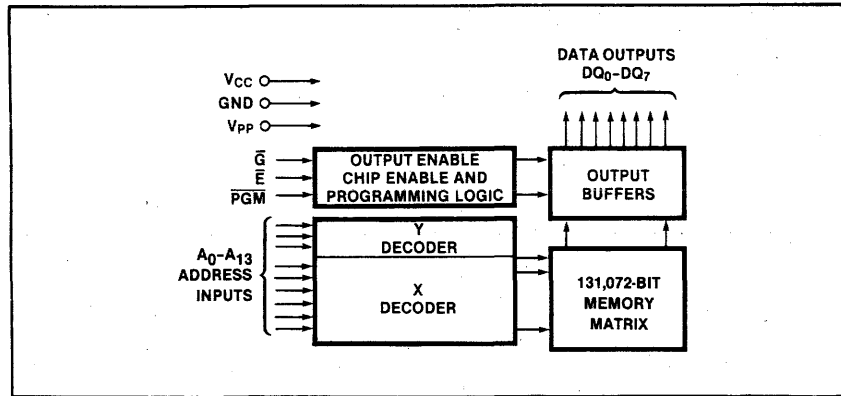
**FEATURES**

- Low CMOS power consumption — 5 $\mu$ W typ. (standby or quiescent active) 25mW typ. (active)
- 250ns access time
- TTL compatible
- Single +5V power supply  $\pm 10\%$
- Industry standard pinout
- Three-state outputs
- Two line controls
- Fast programming

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**PIN NAMES**

A <sub>0</sub> -A <sub>13</sub>	ADDRESSES
$\bar{E}$	Chip Enable
$\bar{G}$	Output Enable
DQ <sub>0</sub> -DQ <sub>7</sub>	Data Inputs/Outputs
PGM	Program Enable

**ABSOLUTE MAXIMUM RATINGS (Note 1)**

PARAMETER	RATING	UNIT
Temperature Range		
Storage	- 65 to + 125	°C
Bias	- 10 to + 80	
Operating Mode	0 to 70	
Program Mode	25 $\pm$ 5	
Applied Voltage with Respect to Ground Potential		
Input	- 0.5 to + 7	V
Output	- 0.5 to + 7	
Supply Voltage with Respect to Ground Potential		
	- 0.5 to + 6	V
	- 0.5 to + 22	

**NOTE**

1. Stresses beyond the limits listed under "Absolute Maximum Ratings" may cause damage to the device. These are stress ratings only. Functional operation at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rated conditions for extended periods of time may affect device reliability.

**131,072-BIT CMOS EPROM (16,384 × 8)**

**27C128**

**Preview**

**DC ELECTRICAL CHARACTERISTICS DURING READ** (Note 2)  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5.0\text{V} \pm 10\%$ , unless otherwise specified,  $V_{PP} = V_{CC}$

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			Min	Typ	Max	
$V_{IL}$	Input Voltage Low		-0.1		0.8	V
$V_{IH}$	Input Voltage High		2.0		$V_{CC} + 1$	V
$V_{OL}$	Output Voltage Low	$I_{OL} = 3.2\text{mA}$	—		0.45	V
$V_{OH}$	Output Voltage High	$I_{OH} = -400\mu\text{A}$	2.4		$V_{CC}$	V
$I_{LI}$	Input Load Current	$V_{IN} = 5.5\text{V}$ or $0.45\text{V}$	—		10	$\mu\text{A}$
$I_{LO}$	Output Leakage	$V_{OUT} = 5.5\text{V}$ or $0.45\text{V}$ , $\bar{E} = V_{CC}$	—		10	$\mu\text{A}$
$I_{CCS1}$	$V_{CC}$ Standby Current	$\bar{E} = V_{CC}$	—	1	100	$\mu\text{A}$
$I_{CCS2}$	$V_{CC}$ Standby Current	$\bar{E} = V_{IH}$	—		1	mA
$I_{CCQ1}$	$V_{CC}$ Active Current (Quiescent)	$\bar{E} = \bar{G} = V_{SS}$ All Addresses at either $V_{SS}$ or $V_{CC}$	—	1	100	$\mu\text{A}$
$I_{CCQ2}$	$V_{CC}$ Active Current (Quiescent)	$\bar{E} = \bar{G} = V_{IL}$ All Addresses at either $V_{IL}$ or $V_{IH}$		1.3	15	mA
$I_{CCA1}$	$V_{CC}$ Active Current (Operating)	$\bar{E} = \bar{G} = V_{SS}$ All addresses changing between $V_{SS}$ and $V_{CC}$	—	5	25	mA
$I_{CCA2}$	$V_{CC}$ Active Current (Operating)	$\bar{E} = \bar{G} = V_{IL}$ All addresses changing between $V_{IL}$ and $V_{IH}$		20	30	mA
$C_{IN}$	Input Capacitance (Note 3)	$V_{IN} = 0\text{V}$ , $f = 1\text{MHz}$		4	6	pF
$C_{OUT}$	Output Capacitance (Note 4)	$V_{OUT} = 0\text{V}$ , $f = 1\text{MHz}$		8	12	pF

- NOTES  
 2. Typical values are for  $T_A = 25^\circ\text{C}$  and nominal power supply voltages.  
 3. Typical values are for  $T_A = 25^\circ\text{C}$  and nominal power supplies.  
 4. These parameters are periodically sampled and are not 100% tested.

**AC ELECTRICAL CHARACTERISTICS DURING READ** (Note 5)  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5\text{V} \pm 10\%$ , unless otherwise specified,  $V_{PP} = V_{CC}$

STANDARD SYMBOL	COMMON SYMBOL	PARAMETER	TEST CONDITIONS (Note 5)	LIMITS						UNIT
				27C128-20		27C128-25		27C128-30		
				Min	Max	Min	Max	Min	Max	
$t_{AVQV}$	$t_{ACC}$	Address Access Time	$\bar{E} = \bar{G} = V_{IL}$		200		250		300	ns
$t_{ELOV}$	$t_{CE}$	Chip Select Delay	$\bar{G} = V_{IL}$		200		250		300	ns
$t_{GLQV}$	$t_{OE}$	Output Enable Delay	$\bar{E} = V_{IL}$ (Note 6)	10	70	10	100	10	150	ns
$t_{EHQZ}$	$t_{DF}$	Chip Deselect Delay (Note 7)	$\bar{E} = V_{IL}$ $t_{DF}$ is measured at $V_{OL} = 0.8$ to tri-state (High-Z) transition	0	60	0	90	0	130	ns
$t_{AXQZ}$	$t_{OH}$	Previous Output Data Holding Time	From Addresses, $\bar{E}$ or $\bar{G}$ whichever occurred first $\bar{E} = \bar{G} = V_{IL}$	0		0		0		ns

- NOTES  
 5. AC TEST CONDITIONS:  
 Output load 2 TTL gates and  $C_L = 100\text{pF}$   
 Input rise and fall times:  $\leq 20\text{ns}$   
 Input pulse levels: 0.8V to 2.2V  
 Timing measurement reference levels:  
 Inputs 1.5V  
 Outputs 0.6V and 2.2V  
 6. G may be delayed up to  $t_{AVQV} - t_{GLQV}$  after the falling edge of  $\bar{E}$  without impact on  $t_{AVQV}$ .  
 7. Specified from  $\bar{G}$  or  $\bar{E}$ , whichever occurs first.

MEMORY Signetics

**131,072-BIT CMOS EPROM (16,384 x 8)**

**27C128**

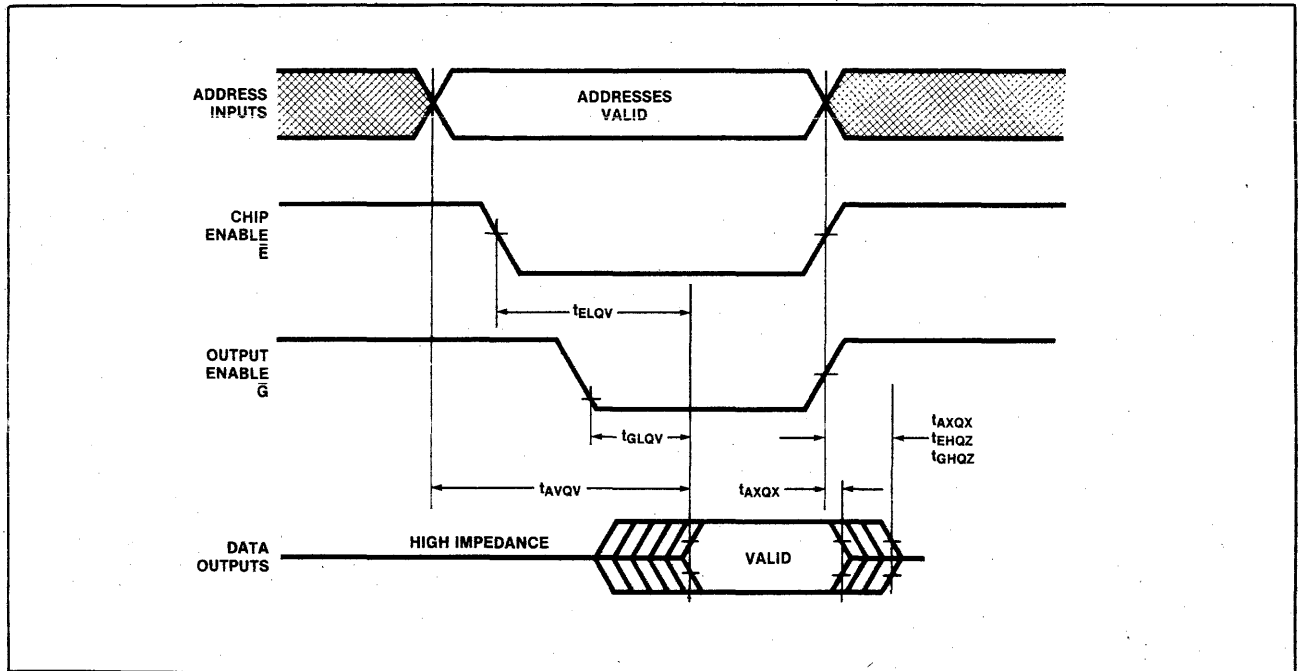
**Preview**

**MODE SELECTION**

MODE/PINS	$\bar{E}$ (20)	$\bar{G}$ (22)	$\overline{PGM}$ (27)	$V_{PP}$ (1)	$V_{CC}$ (28)	DQ <sub>0</sub> -DQ <sub>7</sub> (11-13, 15-19)
Read	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>CC</sub>	V <sub>CC</sub>	DQ <sub>OUT</sub>
Standby	V <sub>IH</sub>	X	X	V <sub>CC</sub>	V <sub>CC</sub>	High Z
Program	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>IL</sub>	V <sub>PP</sub>	V <sub>CC</sub>	DQ <sub>IN</sub>
Program Verify	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>PP</sub>	V <sub>CC</sub>	DQ <sub>OUT</sub>
Program Inhibit	V <sub>IH</sub>	X	X	V <sub>PP</sub>	V <sub>CC</sub>	High Z

X can be either V<sub>OL</sub> or V<sub>OH</sub>.

**TIMING DIAGRAM DURING READ**



**DC ELECTRICAL CHARACTERISTICS DURING PROGRAMMING** (Note 8) T<sub>A</sub> = 25 ± 5°C, V<sub>CC</sub> = 5V ± 5%, V<sub>PP</sub> = 21V ± 0.5V

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			Min	Typ	Max	
V <sub>IL</sub>	Input Voltage — Low		-0.1		0.8	V
V <sub>IH</sub>	Input Voltage — High		2.0		V <sub>CC</sub> + 1	V
V <sub>OL</sub>	Output Voltage During Verify — Low	I <sub>OL</sub> = 3.2mA			0.45	V
V <sub>OH</sub>	Output Voltage During Verify — High	I <sub>OH</sub> = 400μA	2.4			V
I <sub>LI</sub>	Input Load Current	V <sub>IN</sub> = 5.25V or V <sub>IL</sub>			10	μA
I <sub>CCA3</sub>	V <sub>CC</sub> Supply Current — Active			20	30	mA
I <sub>PP2</sub>	V <sub>PP</sub> Supply Current During Programming Pulse	$\bar{E} = \overline{PGM} = V_{IL}$ V <sub>PP</sub> = 21.5V, DQ <sub>0</sub> -DQ <sub>7</sub> = V <sub>IL</sub>			30	mA

NOTE

8. V<sub>CC</sub> must be applied before and removed after V<sub>pp</sub>. V<sub>pp</sub> should never exceed 21.5V (including transients). If  $\overline{PGM} = V_{IL}$ , V<sub>pp</sub> should not be switched.

**131,072-BIT NMOS EPROM (16,384 × 8)**

**27128**

**Preview**

**DESCRIPTION**

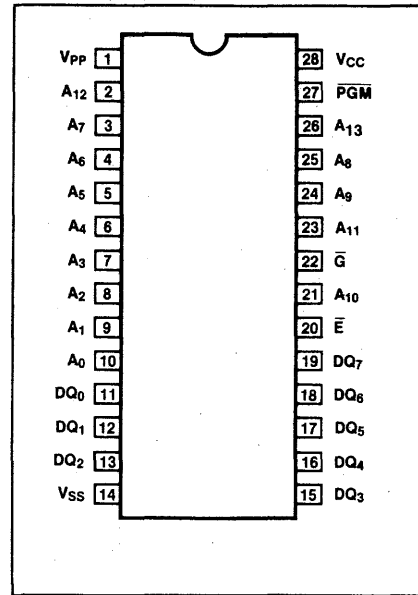
The Signetics 27128 EPROM is an NMOS, 5V only, 131,072-bit electrically programmable and ultraviolet light erasable read only memory, organized as 16,384 words by 8 bits. The 27128 has a separate Output Enable control ( $\bar{G}$ ) from the Chip Enable control ( $\bar{E}$ ), in order to eliminate bus contention in multiple bus systems.

The 27128 has a standby mode to reduce the power dissipation with no increase in access time. The maximum active current is 100mA while the standby current is 40mA from the power supply. This mode is initiated by placing a TTL-high signal on the  $\bar{E}$  input.

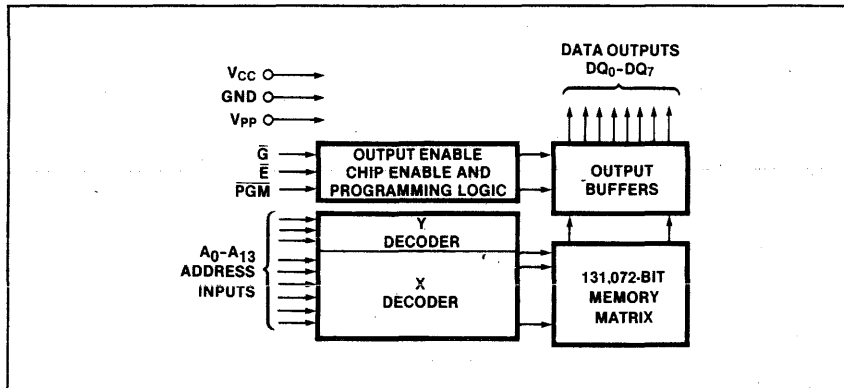
**FEATURES**

- 250ns access time
- TTL compatible
- Single + 5V power supply  $\pm 10\%$
- Industry standard pinout
- Power down mode with 30mA maximum power supply current
- Three-state outputs
- Two line controls

**PIN CONFIGURATION**



**BLOCK DIAGRAM**



**PIN NAMES**

$A_0-A_{13}$	ADDRESSES
$\bar{E}$	Chip Enable
$\bar{G}$	Output Enable
$DQ_0-DQ_7$	Data Inputs/Outputs
$\bar{P}GM$	Program Enable

**DC ELECTRICAL CHARACTERISTICS DURING READ** (Note 1)  $T_A = 0^\circ$  to  $+70^\circ C$ ,  $V_{CC} = 5.0V \pm 10\%$ , unless otherwise specified,  $V_{PP} = V_{CC}$

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			Min	Typ	Max	
$V_{IL}$	Input Voltage — Low		-0.1		0.8	V
$V_{IH}$	Input Voltage — High		2.0		$V_{CC} + 1$	
$V_{OL}$	Output Voltage — Low	$I_{OL} = 3.2mA$			0.45	V
$V_{OH}$	Output Voltage — High	$I_{OH} = -400\mu A$	2.4		$V_{CC}$	
$I_{LI}$	Input Load Current	$V_{IN} = 5.5V$ or $0.45V$			10	$\mu A$
$I_{LO}$	Output Leakage	$V_{OUT} = 5.5V$ or $0.45V$ , $\bar{E} = V_{IH}$			10	$\mu A$
$I_{CC1}$	$V_{CC}$ Supply Current — Standby	$\bar{E} = V_{IH}$		15	30	mA
$I_{CC2}$	$V_{CC}$ Supply Current — Active	$\bar{E} = \bar{G} = V_{IL}$		50	100	mA
$I_{PP1}$	$V_{PP}$ Supply Current — Read Mode	$V_{PP} = 5.25V$			5	mA

NOTE  
1. Typical values are for  $T_A = 25^\circ C$  and nominal power supply voltages.

**131,072-BIT NMOS EPROM (16,384 × 8)**

**27128**

**Preview**

**AC ELECTRICAL CHARACTERISTICS DURING READ** (Note 2)  $T_A = 0^\circ$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5\text{V} \pm 10\%$ , unless otherwise specified,  $V_{PP} = V_{CC}$

STANDARD SYMBOL	COMMON SYMBOL	PARAMETER	TEST CONDITIONS (Note 2)	LIMITS						UNIT
				27128-20		27128-25		27128-30		
				Min	Max	Min	Max	Min	Max	
$t_{AVQV}$	$t_{ACC}$	Address Access Time	$\bar{E} = \bar{G} = V_{IL}$		200		250		300	ns
$t_{ELQV}$	$t_{CE}$	Chip Select Delay	$\bar{G} = V_{IL}$		200		250		300	ns
$t_{GLQV}$	$t_{OE}$	Output Enable Delay	$\bar{E} = V_{IL}$ (Note 3)	10	70	10	100	10	150	ns
$t_{EHQX}$ $t_{GHQX}$	$t_{DF}$	Chip Deselect Delay (Note 4)	$\bar{E} = V_{IL}$ $t_{DF}$ is measured at $V_{OL} = 0.5$ to tri-state (High-Z) transition	0	60	0	90	0	130	ns
$t_{AXQX}$	$t_{OH}$	Previous Output Data Holding Time	From Addresses, $\bar{E}$ or $\bar{G}$ whichever occurred first $\bar{E} = \bar{G} = V_{IL}$	0		0		0		ns

NOTES

2. AC TEST CONDITIONS:

Output load 2 TTL gates and  $C_L = 100\text{pF}$  Timing measurement reference levels:  
 Input rise and fall times:  $\leq 20\text{ns}$  Inputs 1.5V  
 Input pulse levels: 0.8V to 2.2V Outputs 0.6V and 2.2V

3. G may be delayed up to  $t_{AVQV} - t_{GLQV}$  after  $\bar{E}$  without impact on  $t_{AVQV}$ .

4.  $t_{EHQX}$  is specified from  $\bar{G}$  or  $\bar{E}$ , whichever occurs first.

**DC ELECTRICAL CHARACTERISTICS DURING PROGRAMMING** (Note 5)  $T_A = 25 \pm 5^\circ\text{C}$ ,  $V_{CC} = 5\text{V} \pm 5\%$ ,  $V_{PP} = 21\text{V} \pm 0.5\text{V}$

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			Min	Typ	Max	
$V_{IL}$	Input Voltage — Low		-0.1		0.8	V
$V_{IH}$	Input Voltage — High		2.0		$V_{CC} + 1$	
$V_{OL}$	Output Voltage During Verify — Low	$I_{OL} = 3.2\text{mA}$			0.45	V
$V_{OH}$	Output Voltage During Verify — High	$I_{OH} = 400\mu\text{A}$	2.4			
$I_{LI}$	Input Load Current	$V_{IN} = 5.25\text{V}$ or $V_{IL}$			10	$\mu\text{A}$
$I_{CC2}$	$V_{CC}$ Supply Current — Active				100	mA
$I_{PP2}$	$V_{PP}$ Supply Current During Programming Pulse	$\bar{E} = \bar{PGM} = V_{IL}$ $V_{PP} = 21.5\text{V}$ , $DQ_0\text{-}DQ_7 = V_{IL}$			30	mA

NOTE

5.  $V_{CC}$  must be applied before and removed after  $V_{PP}$ .  $V_{PP}$  should never exceed 21.5V (including transients). If  $\bar{PGM} = V_{IL}$ ,  $V_{PP}$  should not be switched.

**AC ELECTRICAL CHARACTERISTICS DURING PROGRAMMING** (Note 7)  $T_A = 25 \pm 5^\circ\text{C}$ ,  $V_{CC} = 5\text{V} \pm 5\%$ ,  $V_{PP} = 21\text{V} \pm 0.5\text{V}$

STANDARD SYMBOL	COMMON SYMBOL	PARAMETER	TEST CONDITIONS (Note 2)	LIMITS			UNIT
				Min	Typ	Max	
$t_{AVPL}$	$t_{AS}$	Address Set-up Time		2			$\mu\text{s}$
$t_{DVPL}$	$t_{DS}$	Data Set-up Time		2			$\mu\text{s}$
$t_{PHDX}$	$t_{DH}$	Data Hold Time		2			$\mu\text{s}$
$t_{VHPL}$	$t_{VPPS}$	$V_{PP}$ Set-up Time		2			$\mu\text{s}$
$t_{ELPL}$	$t_{CES}$	$\bar{E}$ Set-up Time		2			$\mu\text{s}$
$t_{PLPH}$	$t_{PW}$	$\bar{PGM}$ Pulse Width During Programming		45	50	55	ms
$t_{GHPL}$	$t_{OES}$	$\bar{G}$ Set-up Time		2			$\mu\text{s}$
$t_{GHQX}$ $t_{EHQZ}$	$t_{DF}$	Chip Deselect Delay (Note 6)	$\bar{E} = V_{IL}$	0		130	ns
$t_{PHAX}$	$t_{AH}$	Address Hold Time			0		ns

NOTES

6.  $t_{GHQX}$  is specified from  $\bar{G}$  or  $\bar{E}$ , whichever occurs first.

7. The input reference level is 1V for  $V_{IL}$  and 2V for  $V_{IH}$ .

# 65,536-BIT STATIC MOS ROM (8192 x 8)

2364-15/20

## DESCRIPTION

This ROM is designed for memory applications where high performance, large bit storage, and simple interfacing are important design objectives.

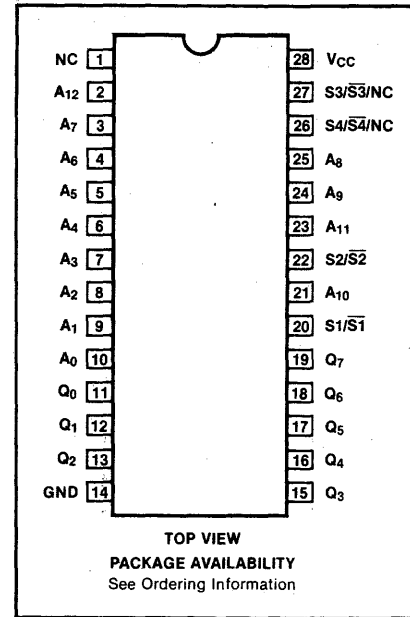
The four Chip Select inputs are programmable. Active-high or active-low level Chip Select input can be defined by the designer and the desired chip select logic level is fixed during the masking process. The programmable Chip Select input, as well as OR-tie compatibility on the outputs, facilitates easy memory expansion.

The 2364 Read-Only Memory is fabricated with n-channel silicon gate technology. This technology provides the designer with high performance, easy-to-use MOS circuits.

## FEATURES

- Fast access time
- Industry standard pinout — JEDEC approved
- Completely TTL compatible
- One +5V ± 10% power supply
- Low power dissipation
- Three-state output — OR-tie capability
- Four programmable Chip Select inputs for easy memory expansion or no connection option
- Inputs protected — all inputs have protection against static charge
- 2764 EPROM compatible

## PIN CONFIGURATION



## ABSOLUTE MAXIMUM RATINGS (Note 1)

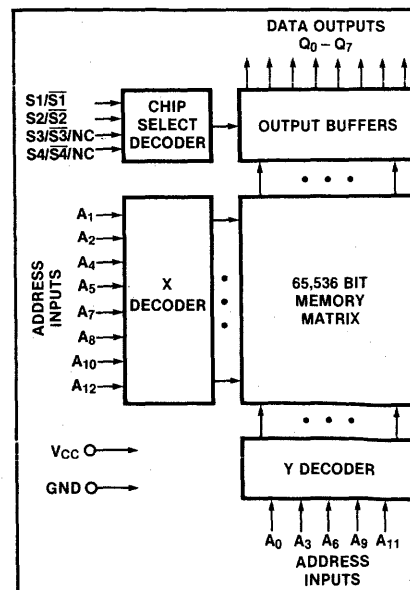
PARAMETER	RATING	UNIT
Operating Temperature 2364-XX	0 to +70	°C
2364E-XX	-40 to +85	°C
Storage Temperature	-65 to +150	°C
Voltage on Any Pin Relative to Ground	-2 to +7	V
Power Dissipation	1	Watt
Electrostatic Discharge Rating Relative to Ground (Note 2)	± 2000	V

- NOTES
- Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these any other conditions above those indicated in the operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.
  - Test condition. Discharge 100pF through 1500 ohms, MIL-STD 883, method 3015.1.

## PIN NAMES

A <sub>0</sub> -A <sub>12</sub>	Address Inputs
S1-S4	Chip Selects
Q <sub>0</sub> -Q <sub>7</sub>	Data Outputs
V <sub>CC</sub>	Power
GND	Ground
NC	Not Connected

## BLOCK DIAGRAM



## ORDERING INFORMATION

TEMPERATURE RANGE	PACKAGE TYPE	ORDER NUMBER BY ACCESS TIME	
		150ns	200ns
0°C to +70°C	Plastic	2364-15N	2364-20N
0°C to +70°C	Cerdip	2364-15F	2364-20F
0°C to +70°C	Ceramic	2364-15I	2364-20I
-40°C to +85°C	Plastic	2364E-15N	2364E-20N
-40°C to +85°C	Cerdip	2364E-15F	2364E-20F
-40°C to +85°C	Ceramic	2364E-15I	2364E-20I



**65,536-BIT STATIC MOS ROM (8192 × 8)**

**2364-15/20**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5.0V \pm 10\%$

PARAMETER	TEST CONDITIONS	2364E-15		2364E-20		2364-15		2364-20		UNIT
		Min	Max	Min	Max	Min	Max	Min	Max	
$V_{IL}$ $V_{IH}$	Input voltage Low (Note 3) High	-2.0 2.0	0.8 $V_{CC}$	-2.0 2.0	0.8 $V_{CC}$	-2.0 2.0	0.8 $V_{CC}$	-2.0 2.0	0.8 $V_{CC}$	V
$V_{OL}$ $V_{OH}$	Output voltage Low High		0.4 $V_{CC}$		0.4 $V_{CC}$		0.4 $V_{CC}$		0.4 $V_{CC}$	V
$I_{LI}$	Input load current		1		1		1		1	$\mu A$
$I_{LO}$	Output leakage	Chip deselected. $V_{OUT} = +0.4V$ to $V_{CC}$		5		5		5		$\mu A$
$I_{CC}$	Supply current	Chip deselected. $V_{CC} = 5.5V$ , $V_{IN} = V_{CC}$		120		110		90		mA
$C_{IN}$ $C_{OUT}$	Capacitance (Note 4) Input Output	$T_A = 25^\circ C$ , $f = 1MHz$ , all pins except pin under test tied to ground		7 10		7 10		7 10		pF

NOTES

- 3. Input levels that swing more negative than -2.0V may alter AC electrical characteristics.
- 4. This parameter is periodically sampled and is not 100% tested.

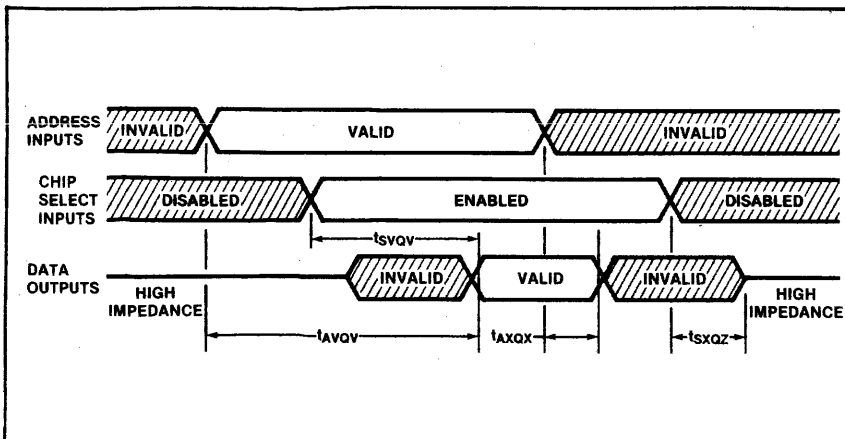
**AC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5.0V \pm 10\%$

PARAMETER	-15		-20		UNIT	
	Min	Max	Min	Max		
$t_{AVQV}$	Address access time			150	200	ns
$t_{SVQV}$ (Note 5)	Chip select delay			100	100	ns
$t_{SXQZ}$ (Note 6)	Chip deselect delay			75	100	ns
$t_{AXQX}$	Previous data valid after address change delay		0		0	ns

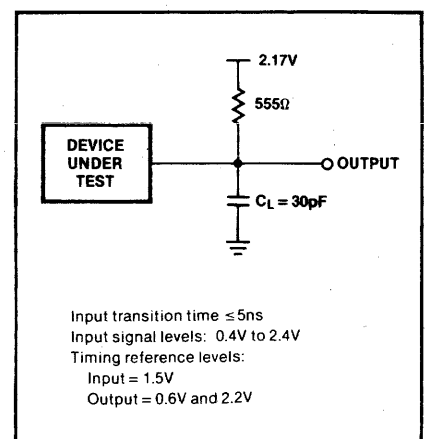
NOTES

- 5.  $t_{SVQV}$  is at value indicated, provided the valid address leads the chip select by  $(t_{AVQV} - t_{SVQV})$  ns or more.
- 6.  $t_{SXQZ}$  is measured at  $V_{OL} = 0.5V$  on the "0" to High-Z state transition, with AC test load connected to the output.

**TIMING DIAGRAM**



**AC TEST LOAD**



**65,536-BIT STATIC MOS ROM (8192 × 8)**

**2664A-15/20**

**DESCRIPTION**

This ROM is designed for memory applications where high performance, large bit storage, and simple interfacing are important design objectives.

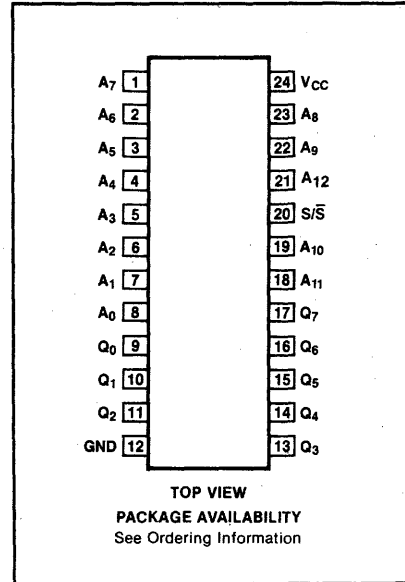
The Chip Select input is programmable. Active-high or active-low level Chip Select input can be defined by the designer and the desired chip select logic level is fixed during the masking process. The programmable Chip Select input, as well as OR-tie compatibility on the outputs, facilitates easy memory expansion.

The 2664A Read-Only Memory is fabricated with n-channel silicon gate technology. This technology provides the designer with high performance, easy-to-use MOS circuits.

**FEATURES**

- **Fast access time**
- **Low power dissipation**
- **Industry standard pinout — JEDEC approved**
- **Completely TTL compatible**
- **One +5V ± 10% power supply**
- **Three-state output — OR-tie capability**
- **One programmable chip select input for easy memory expansion or no connection option**
- **Inputs protected — all inputs have protection against static charge**
- **2764 EPROM compatible**

**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS (Note 1)**

PARAMETER	RATING	UNIT
Operating Temperature 2664A-XX	0 to +70	°C
2664AE-XX	-40 to +85	°C
Storage Temperature	-65 to +150	°C
Voltage on Any Pin Relative to Ground	-2 to +7	V
Power Dissipation	1	Watt
Electrostatic Discharge Rating Relative to Ground (Note 2)	± 2000	V

**NOTES**

1. Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these any other conditions above those indicated in the operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.
2. Test condition. Discharge 100pF through 1500 ohms, MIL-STD 883, method 3015.1.

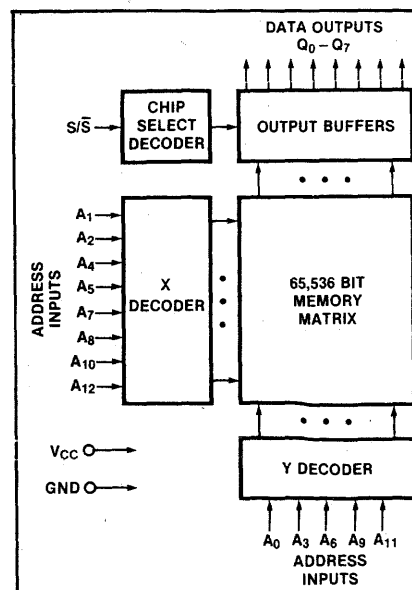
**PIN NAMES**

A <sub>0</sub> -A <sub>12</sub>	Address Inputs
S	Chip Select
Q <sub>0</sub> -Q <sub>7</sub>	Data Outputs
V <sub>CC</sub>	Power
GND	Ground

**ORDERING INFORMATION**

TEMPERATURE RANGE	PACKAGE TYPE	ORDER NUMBER BY ACCESS TIME	
		150ns	200ns
0°C to +70°C	Plastic	2664A-15N	2664A-20N
-40°C to +85°C	Plastic	2664AE-15N	2664AE-20N
-40°C to +85°C	Cerdip	2664AE-15F	2664AE-20F
-40°C to +85°C	Ceramic	2664AE-15I	2664AE-20I

**BLOCK DIAGRAM**



MEMORY Signetics

# 65,536-BIT STATIC MOS ROM (8192 × 8)

2664A-15/20

## DC ELECTRICAL CHARACTERISTICS $V_{CC} = 5.0V \pm 10\%$

PARAMETER	TEST CONDITIONS	2664AE-15		2664AE-20		2664A-15		2664A-20		UNIT
		Min	Max	Min	Max	Min	Max	Min	Max	
$V_{IL}$ Input voltage Low (Note 3)		-2.0	0.8	-2.0	0.8	-2.0	0.8	-2.0	0.8	V
$V_{IH}$ Input voltage High		2.0	$V_{CC}$	2.0	$V_{CC}$	2.0	$V_{CC}$	2.0	$V_{CC}$	V
$V_{OL}$ Output voltage Low	$I_{OL} = 3.2mA$ $I_{OH} = -400\mu A$		0.4		0.4		0.4		0.4	V
$V_{OH}$ Output voltage High		2.4	$V_{CC}$	2.4	$V_{CC}$	2.4	$V_{CC}$	2.4	$V_{CC}$	V
$I_{LI}$ Input load current	$0V \leq V_{IN} \leq 5.5V$		1		1		1		1	$\mu A$
$I_{LO}$ Output leakage	Chip deselected. $V_{OUT} = +0.4V$ to $V_{CC}$		5		5		5		5	$\mu A$
$I_{CC}$ Supply current	Chip deselected. $V_{CC} = 5.5V$ , $V_{IN} = V_{CC}$		120		110		110		90	mA
Capacitance (Note 4)	$T_A = 25^\circ C$ , $f = 1MHz$ , all pins except pin under test tied to ground		7		7		7		7	pF
$C_{IN}$ Input			10		10		10		10	
$C_{OUT}$ Output										

NOTES

- 3. Input levels that swing more negative than -2.0V may alter AC electrical characteristics.
- 4. This parameter is periodically sampled and is not 100% tested.

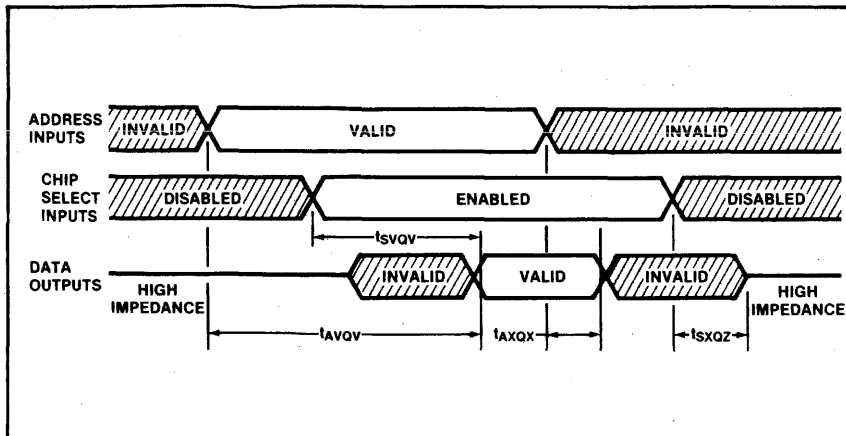
## AC ELECTRICAL CHARACTERISTICS $V_{CC} = 5.0V \pm 10\%$

PARAMETER	-15		-20		UNIT
	Min	Max	Min	Max	
$t_{AVQV}$ Address access time		150		200	ns
$t_{SVQV}$ (Note 5) Chip select delay		100		100	ns
$t_{SXQZ}$ (Note 6) Chip deselect delay		75		100	ns
$t_{AXQX}$ Previous data valid after address change delay	0		0		ns

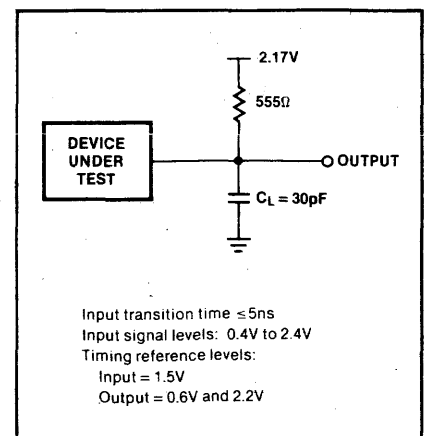
NOTES

- 5.  $t_{SVQV}$  is at value indicated, provided the valid address leads the chip select by  $(t_{AVQV} - t_{SVQV})$  ns or more.
- 6.  $t_{SXQZ}$  is measured at  $V_{OL} = 0.5V$  on the "0" to High-Z state transition, with AC test load connected to the output.

### TIMING DIAGRAM



### AC TEST LOAD



# 131,072-BIT STATIC MOS ROM (16,384 × 8)

23128-15/20

## DESCRIPTION

This ROM is designed for memory applications where high performance, large bit storage, and simple interfacing are important design objectives.

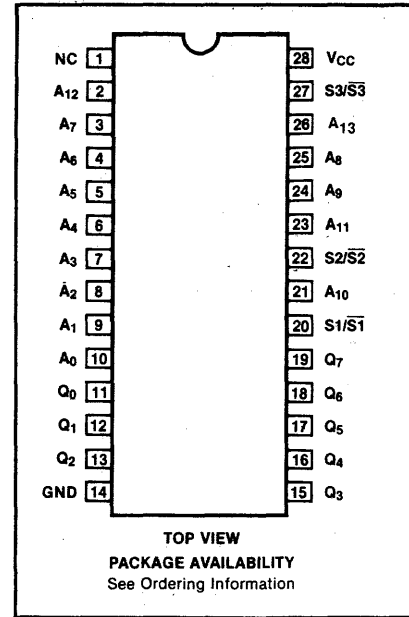
The three Chip Select inputs are programmable. Active-high or active-low level Chip Select input can be defined by the designer and the desired chip select logic level is fixed during the masking process. The programmable Chip Select input, as well as OR-tie compatibility on the outputs, facilitates easy memory expansion.

The 23128 Read-Only Memory is fabricated with n-channel silicon gate technology. This technology provides the designer with high performance, easy-to-use MOS circuits.

## FEATURES

- Fast access time
- Industry standard pinout — JEDEC approved
- Completely TTL compatible
- One +5V ± 10% power supply
- Low power dissipation
- Three-state output — OR-tie capability
- Four programmable Chip Select inputs for easy memory expansion or no connection option
- Inputs protected — all inputs have protection against static charge
- 27128 EPROM compatible

## PIN CONFIGURATION



## ABSOLUTE MAXIMUM RATINGS (Note 1)

PARAMETER	RATING	UNIT
Operating Temperature 23128-XX	0 to +70	°C
23128E-XX	-40 to +85	°C
Storage Temperature	-65 to +150	°C
Voltage on Any Pin Relative to Ground	-2 to +7	V
Power Dissipation	1	Watt
Electrostatic Discharge Rating Relative to Ground (Note 2)	± 2000	V

### NOTES

- Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these any any other conditions above those indicated in the operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.
- Test condition. Discharge 100pF through 1500 ohms, MIL-STD 883, method 3015.1.

## PIN NAMES

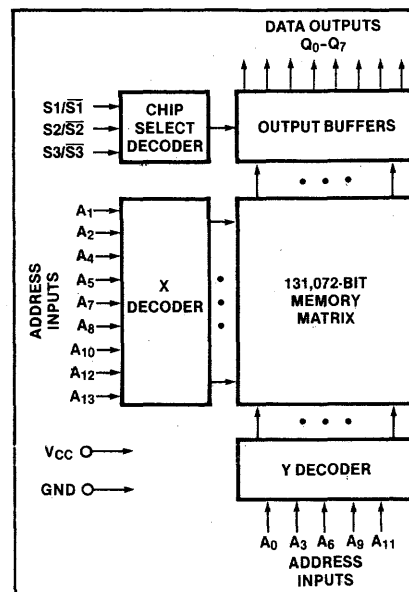
A <sub>0</sub> -A <sub>13</sub>	Address Inputs
S1-S3	Chip Selects
Q <sub>0</sub> -Q <sub>7</sub>	Data Outputs
V <sub>CC</sub>	Power
GND	Ground

## ORDERING INFORMATION

TEMPERATURE RANGE	PACKAGE TYPE	ORDER NUMBER BY ACCESS TIME	
		150ns*	200ns
0°C to +70°C	Plastic	23128-15N	23128-20N
0°C to +70°C	Ceramic	23128-15I	23128-20I
-40°C to +85°C	Plastic	23128E-15N	23128E-20N
-40°C to +85°C	Ceramic	23128E-15I	23128E-20I

\*Preliminary

## BLOCK DIAGRAM



**131,072-BIT STATIC MOS ROM (16,384 × 8)**

**23128-15/20**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5.0V \pm 10\%$

PARAMETER	TEST CONDITIONS	23128E-15*		23128E-20		23128-15*		23128-20		UNIT
		Min	Max	Min	Max	Min	Max	Min	Max	
$V_{IL}$ Input voltage Low (Note 3)		-2.0	0.8	-2.0	0.8	-2.0	0.8	-2.0	0.8	V
$V_{IH}$ Input voltage High		2.0	$V_{CC}$	2.0	$V_{CC}$	2.0	$V_{CC}$	2.0	$V_{CC}$	V
$V_{OL}$ Output voltage Low	$I_{OL} = 3.2mA$ $I_{OH} = -400\mu A$		0.4		0.4		0.4		0.4	V
$V_{OH}$ Output voltage High		2.4	$V_{CC}$	2.4	$V_{CC}$	2.4	$V_{CC}$	2.4	$V_{CC}$	V
$I_{LI}$ Input load current	$0V \leq V_{IN} \leq 5.5V$		1		1		1		1	$\mu A$
$I_{LO}$ Output leakage	Chip deselected. $V_{OUT} = +0.4V$ to $V_{CC}$		5		5		5		5	$\mu A$
$I_{CC}$ Supply current	Chip deselected. $V_{CC} = 5.5V$ , $V_{IN} = V_{CC}$		120		120		110		100	mA
Capacitance (Note 4)	$T_A = 25^\circ C$ , $f = 1MHz$ , all pins except pin under test tied to ground		7		7		7		7	pF
$C_{IN}$ Input				10		10		10		10
$C_{OUT}$ Output										pF

NOTES

- 3. Input levels that swing more negative than -2.0V may alter AC electrical characteristics.
- 4. This parameter is periodically sampled and is not 100% tested.

**AC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5.0V \pm 10\%$

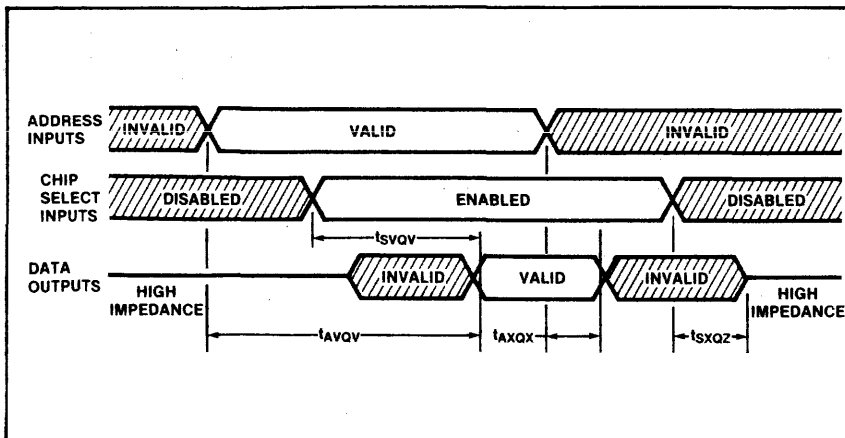
PARAMETER	-15*		-20		UNIT
	Min	Max	Min	Max	
$t_{AVQV}$ Address access time		150		200	ns
$t_{SVQV}$ (Note 5) Chip select delay		100		100	ns
$t_{SXQZ}$ (Note 6) Chip deselect delay		75		100	ns
$t_{AXQX}$ Previous data valid after address change delay	0		0		ns

NOTES

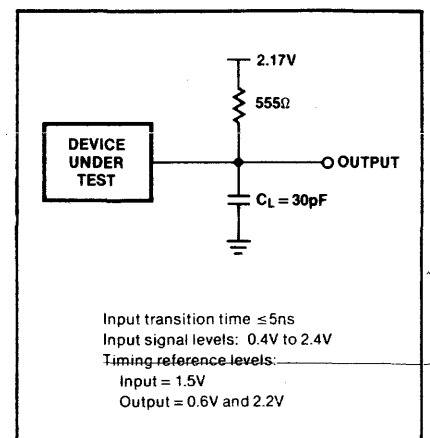
- 5.  $t_{SVQV}$  is at value indicated, provided the valid address leads the chip select by  $(t_{AVQV} - t_{SVQV})$  ns or more.
- 6.  $t_{SXQZ}$  is measured at  $V_{OL} = 0.5V$  on the "0" to High-Z state transition, with AC test load connected to the output.

\* Preliminary

**TIMING DIAGRAM**



**AC TEST LOAD**



# 131,072-BIT STATIC MOS ROM (16,384 x 8)

# 23128/23128E/23128M

## DESCRIPTION

This ROM is designed for memory applications where high performance, large bit storage, simple interfacing and wide operating temperature range are important design objectives.

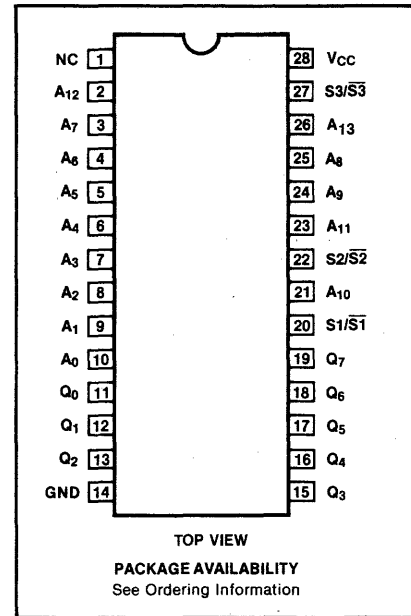
The Chip Select inputs are programmable. Active-high or active-low level Chip Select input can be defined by the designer and the desired chip select logic level is fixed during the masking process. The programmable Chip Select input, as well as OR-tie compatibility on the outputs, facilitates easy memory expansion.

The 23128 Read-Only Memory is fabricated with n-channel silicon gate technology. This technology provides the designer with high performance, easy-to-use MOS circuits.

## FEATURES

- **Fast access time**
  - 250ns max. 23128-25
  - 300ns max. 23128M
- **Industry standard pinout — JEDEC approved**
- **Completely TTL compatible**
- **Single +5V ± 10% power supply**
- **Low power dissipation**
- **Three-state output — OR-tie capability**
- **Three programmable Chip Select inputs for easy memory expansion**
- **Inputs protected — all inputs have protection against static charge**
- **27128 compatible**

## PIN CONFIGURATION



## ABSOLUTE MAXIMUM RATINGS (Note 1)

PARAMETER	RATING	UNIT	
Operating Temperature	23128-XX	0 to +70	°C
	23128E-XX	-40 to +85	°C
	23128M-XX	-55 to +125	°C
Storage Temperature		-65 to +150	°C
Voltage on Any Pin Relative to Ground	-2 to +7	V	
Power Dissipation	1	Watt	
Electrostatic Discharge Rating Relative to Ground (Note 2)	± 2000	V	

### NOTES

1. Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these or any other conditions above those indicated in the operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.

2. Test condition: Discharge 100pF through 1500ohm, MIL-STD-883, method 3015.1.

## PIN NAMES

A <sub>0</sub> -A <sub>13</sub>	Address Inputs
S <sub>1</sub> -S <sub>3</sub>	Chip Selects
Q <sub>0</sub> -Q <sub>7</sub>	Data Outputs
V <sub>CC</sub>	Power
GND	Ground

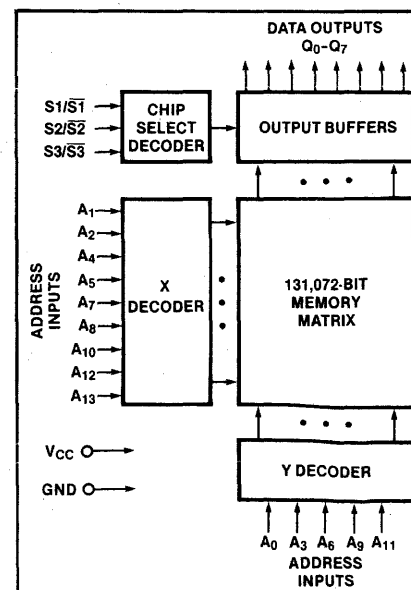
## ORDERING INFORMATION

TEMPERATURE RANGE	PACKAGE TYPE	ORDER NUMBER BY ACCESS TIME		
		250ns	300ns	450ns
0°C to +70°C	Plastic	23128-25N	23128-30N	23128-45N
0°C to +70°C	Ceramic	23128-25I	23128-30I	23128-45I
-40°C to +85°C	Plastic	23128E-25N	23128E-30N	23128E-45N
-40°C to +85°C	Ceramic	23128E-25I	23128E-30I	23128E-45I
-55°C to +125°C (Note 3)	Ceramic	23128M-25	23128M-30I	23128M-45I

### NOTE

3. Also available processed to MIL-STD-883, Method 5004, Class B.

## BLOCK DIAGRAM



# 131,072-BIT STATIC MOS ROM (16,384 × 8)

# 23128/23128E/23128M

## DC ELECTRICAL CHARACTERISTICS $V_{CC} = 5.0V \pm 10\%$

PARAMETER	TEST CONDITIONS	23128E/M		23128-25		23128-30/45		UNIT
		Min	Max	Min	Max	Min	Max	
$V_{IL}$ Input voltage Low (Note 4)	$V_{IH}$ High	-2.0	0.8	-2.0	0.8	-2.0	0.8	V
		2.0	$V_{CC}$	2.0	$V_{CC}$	2.0	$V_{CC}$	
$V_{OL}$ Output voltage Low	$V_{OH}$ High	$I_{OL} = 3.2mA$						V
		$I_{OH} = -400\mu A$		2.4	$V_{CC}$	2.4	$V_{CC}$	
$I_{LI}$ Input load current	$0V \leq V_{IN} \leq 5.5V$		1		1		1	$\mu A$
$I_{LO}$ Output leakage	Chip deselected. $V_{OUT} = +0.4V$ to $V_{CC}$		5		5		5	$\mu A$
$I_{CC}$ Supply current	Chip deselected. $V_{CC} = 5.5V$ , $V_{IN} = V_{CC}$		120		100		90	mA
Capacitance (Note 5)								pF
$C_{IN}$ Input	$T_A = 25^\circ C$ , $f = 1MHz$ , all pins except pin under test tied to ground	7		7		7		
$C_{OUT}$ Output		10		10		10		

NOTES

- 4. Input levels that swing more negative than -2.0V may alter AC electrical characteristics.
- 5. This parameter is periodically sampled and is not 100% tested.

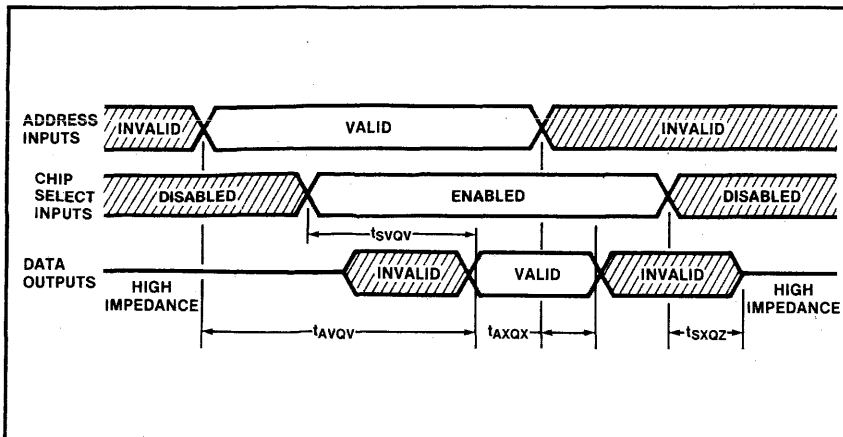
## AC ELECTRICAL CHARACTERISTICS $V_{CC} = 5.0V \pm 10\%$

PARAMETER	-25		-30		-45		UNIT
	Min	Max	Min	Max	Min	Max	
$t_{AVQV}$ Address access time		250		300		450	ns
$t_{SVQV}$ (Note 6) Chip select delay		120		150		150	ns
$t_{SXQZ}$ (Note 7) Chip deselect delay		100		100		100	ns
$t_{AXQX}$ Previous data valid after address change delay	0		0		0		ns

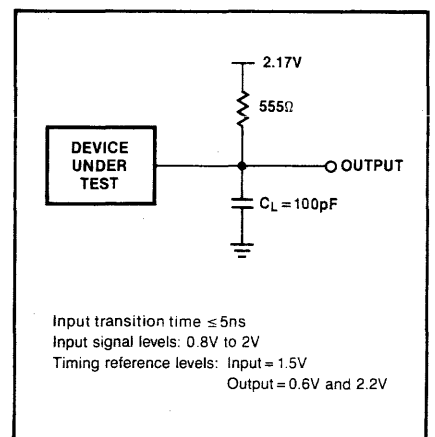
NOTES

- 6.  $t_{SVQV}$  is at value indicated, provided the valid address leads the chip select by ( $t_{AVQV} - t_{SVQV}$ ) ns or more.
- 7.  $t_{SXQZ}$  is measured at  $V_{OL} = 0.5V$  on the "0" to High-Z state transition, with AC TEST LOAD as shown below.

### TIMING DIAGRAM



### AC TEST LOAD



# 262,144-BIT STATIC MOS ROM (32,768 × 8)

23256A

**Preliminary**

**DESCRIPTION**

The 23256A is a 262,144-bit read-only memory organized as 32,768 words by 8 bits. The pinout conforms to JEDEC standards for byte-wide memories.

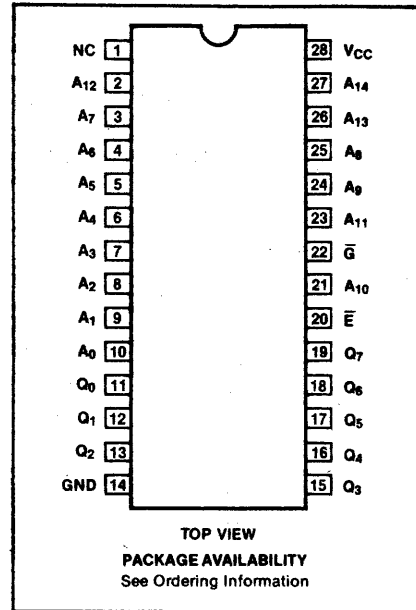
The 23256A is a fully static device; it goes into low power, standby mode when pin 20 ( $\bar{E}$ ), is held high. Either  $\bar{G}$  or  $\bar{E}$  may be used for output control to eliminate bus contention.

The 23256A is fabricated using high performance n-channel silicon gate MOS technology.

**FEATURES**

- Completely TTL compatible, 2 TTL loads
- Single +5V ± 10% power supply
- Three-state outputs
- Fast access time — 200ns
- Fully static
- Low standby power
- JEDEC approved byte-wide pinout
- Inputs and outputs protected against static charge

**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS (Note 1)**

PARAMETER	RATING	UNIT
Operating Temperature	0 to +70	°C
Storage Temperature	-65 to +150	°C
Voltage on Any Pin Relative to Ground	-2 to +7	V
Power Dissipation	1	Watt
Electrostatic Discharge Rating Relative to Ground (Note 2)	± 2000	V

- NOTES**
1. Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these or any other conditions above those indicated in the operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.
  2. Test condition. Discharge 100pF through 1500 ohms, MIL-STD 883, method 3015.1.

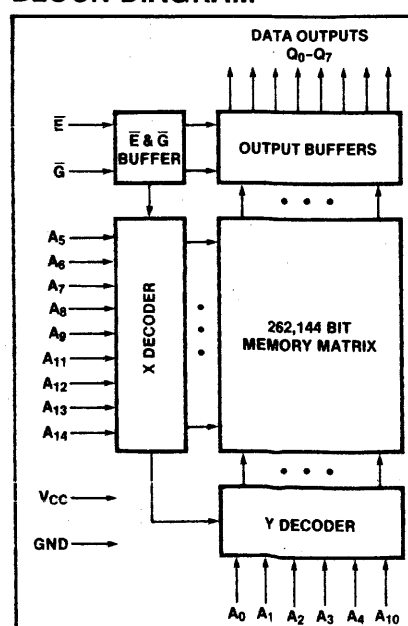
**PIN NAMES**

A <sub>0</sub> -A <sub>14</sub>	Address Inputs
$\bar{E}$	Chip Enable
$\bar{G}$	Output Enable
Q <sub>0</sub> -Q <sub>7</sub>	Data Outputs
V <sub>CC</sub>	Power
GND	Ground
NC	Not Connected

**ORDERING INFORMATION**

TEMPERATURE RANGE	PACKAGE TYPE	ORDER NUMBER BY ACCESS TIME			
		200ns	250ns	300ns	450ns
0°C to +70°C	Plastic	23256A-20N	23256A-25N	23256A-30N	23256A-45N

**BLOCK DIAGRAM**



MEMORY Signetics



**262,144-BIT STATIC MOS ROM (32,768 × 8)**

**23256A**

**Preliminary**

**DC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5.0V \pm 10\%$

PARAMETER	TEST CONDITIONS	Min	Max	UNIT
$V_{IL}$ $V_{IH}$	Input voltage Low (Note 3)	-2.0	0.8	V
	High	2.0	$V_{CC}$	V
$V_{OL}$ $V_{OH}$	Output voltage Low	2.4	0.4	V
	High		$V_{CC}$	V
$I_{LI}$	Input load current	$0V \leq V_{IN} \leq 5.5V$		$\mu A$
$I_{LO}$	Output leakage		5	$\mu A$
$I_{CC}$ $I_{CCZ}$	Supply current Active	Output unloaded $\bar{E} = V_{IH}$		90 mA
	Standby		10	mA
$C_{IN}$ $C_{OUT}$	Capacitance (Note 4) Input	$T_A = 25^\circ C, f = 1MHz$		7 pF
	Output	All pins except pin under test tied to ground		12 pF

NOTES

- 3. Input levels that swing more negative than -2.0V may alter AC electrical characteristics.
- 4. This parameter is periodically sampled and is not 100% tested.

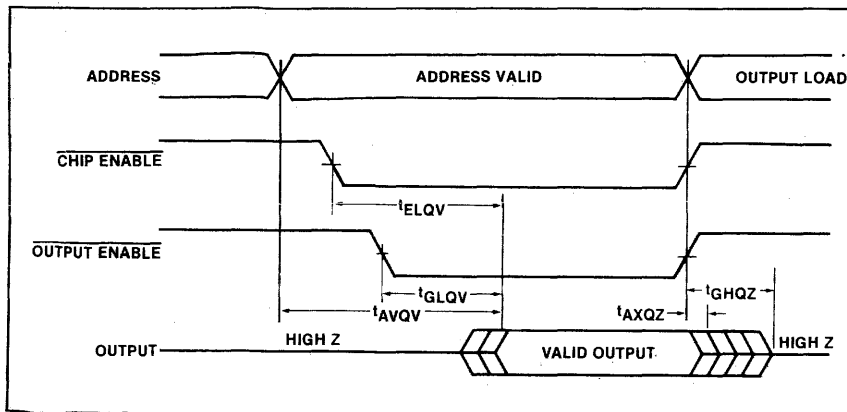
**AC ELECTRICAL CHARACTERISTICS**  $V_{CC} = 5.0V \pm 10\%$

PARAMETER	-20		-25		-30		-45		UNIT
	Min	Max	Min	Max	Min	Max	Min	Max	
$t_{AVQV}$	200		250		300		450		ns
$t_{ELQV}$	200		250		300		450		ns
$t_{GLQV}$	100		120		150		180		ns
$t_{EHQZ}$ $t_{GHQZ}$	100		100		120		150		ns
$t_{AXQZ}$	0		0		0		0		ns

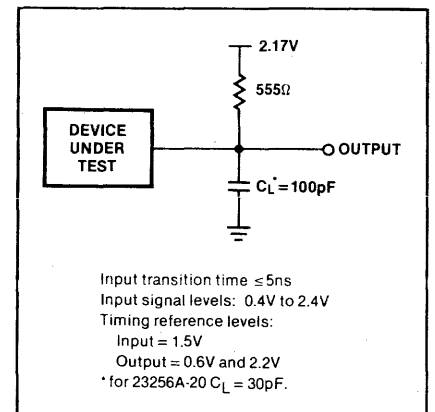
NOTES

- 5.  $t_{ELQV}$  is at value indicated provided the valid address leads output enable by  $(t_{AVQV} - t_{GLQV})$  ns.
- 6.  $t_{EHQZ}$  and  $t_{GHQZ}$  are measured at  $V_{OL} = 0.5V$  on the "0" to High-Z state transition with load as shown below connected to output.

**TIMING DIAGRAM**



**AC TEST LOAD**



**65,536-BIT STATIC MOS ROM (TWO 4096 × 8 BANKS)**

**26S64-30/45**

**DESCRIPTION**

The 26S64 is a mask-programmable, 8192 words by 8 bits Read-Only Memory, partitioned into two banks of 4096 words by 8 bits. Each half can be accessed by a 12-bit address, A<sub>0</sub> through A<sub>11</sub>, and switching from one bank to the other is effected by a pair of bank-select addresses, FF8 and FF9. The address FF8 accesses the bank containing locations 0000 through 0FFF, and FF9 accesses the bank containing locations 1000 through 1FFF. The bank-select locations, 1FF8 and 0FF9, must be programmed with a no-operation (NOP) instruction.

The 26S64 has two user-programmable chip selects, along with three-state outputs, allowing easy memory expansion. It is suitable for applications where 12 ad-

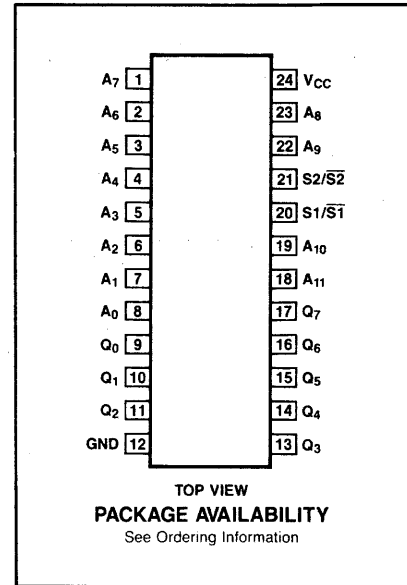
dress lines must be used to access 8192 words of memory, the 13th address bit is internally generated and latched, in response to bank select address FF8 or FF9.

The 26S64 is manufactured with n-channel silicon gate MOS technology, and all inputs and outputs are TTL compatible.

**FEATURES**

- Fast access time
- Organized as two 4K × 8 banks
- Access 8K bytes with 4K system address capability
- Industry standard pinout
- Low power dissipation
- TTL compatible inputs and outputs
- Single +5V ± 10% power supply
- Three-state outputs — OR-tie capability
- Electrostatic discharge protection

**PIN CONFIGURATION**



**ABSOLUTE MAXIMUM RATINGS (Note 1)**

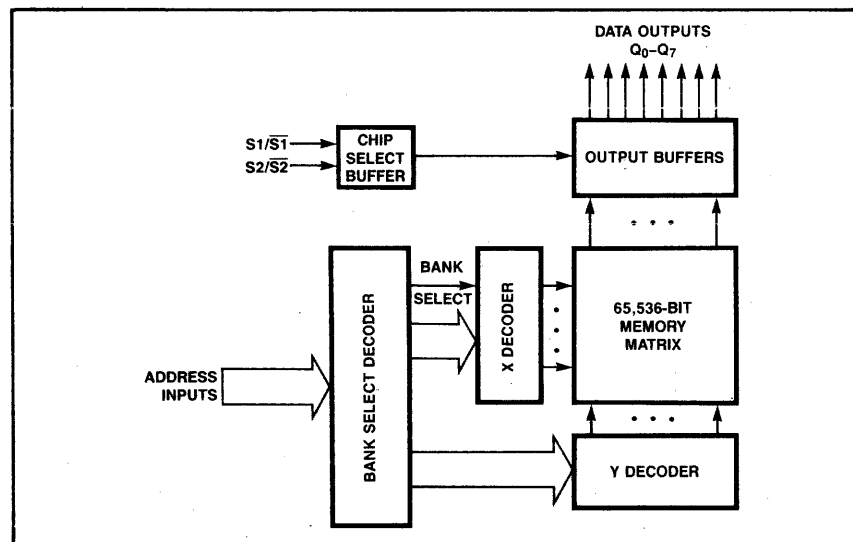
PARAMETER	RATING	UNIT
Operating Temperature	0 to +70	°C
Storage Temperature	-65 to +150	°C
Voltage on Any Pin Relative to Ground	-2 to +7	V
Power Dissipation	1	Watt
Electrostatic Discharge Rating Relative to Ground (Note 2)	± 2000	V

- NOTES**
1. Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these any any other conditions above those indicated in the operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.
  2. Test condition. Discharge 100pF through 1500 ohms. MIL-STD 883, method 3015.1.

**PIN NAMES**

A <sub>0</sub> -A <sub>11</sub>	Address inputs
Q <sub>0</sub> -Q <sub>7</sub>	Data outputs
S1/S2	Chip selects
V <sub>CC</sub> /GND	Power/ground

**BLOCK DIAGRAM**



**ORDERING INFORMATION**

TEMPERATURE RANGE	PACKAGE TYPE	ORDER NUMBER BY ACCESS TIME	
		300ns	450ns
0°C to +70°C	Plastic	26S64-30N	26S64-45N
0°C to +70°C	Cerdip	26S64-30F	26S64-45F

# 65,536-BIT STATIC MOS ROM (TWO 4096 × 8 BANKS)

26S64-30/45

## DC ELECTRICAL CHARACTERISTICS $V_{CC} = 5.0V \pm 10\%$

PARAMETER	TEST CONDITIONS	26S64		UNIT
		Min	Max	
$V_{IL}$ $V_{IH}$	Input voltage Low (Note 3) High	-2.0 2.0	0.8 $V_{CC}$	V
$V_{OL}$ $V_{OH}$	Output voltage Low High		0.4 $V_{CC}$	V
$I_{LI}$	Input load current $0V \leq V_{IN} \leq 5.5V$			$\mu A$
$I_{LO}$	Output leakage Chip deselected. $V_{OUT} = +0.4V$ to $V_{CC}$		5	$\mu A$
$I_{CC}$	Supply current Chip deselected. $V_{CC} = 5.5V$ , $V_{IN} = V_{CC}$		100	mA
$C_{IN}$ $C_{OUT}$	Capacitance (Note 4) Input Output $T_A = 25^\circ C$ , $f = 1MHz$ , all pins except pin under test tied to ground		7 10	pF

NOTES

- 3. Input levels that swing more negative than -2.0V may alter AC electrical characteristics.
- 4. This parameter is periodically sampled and is not 100% tested.

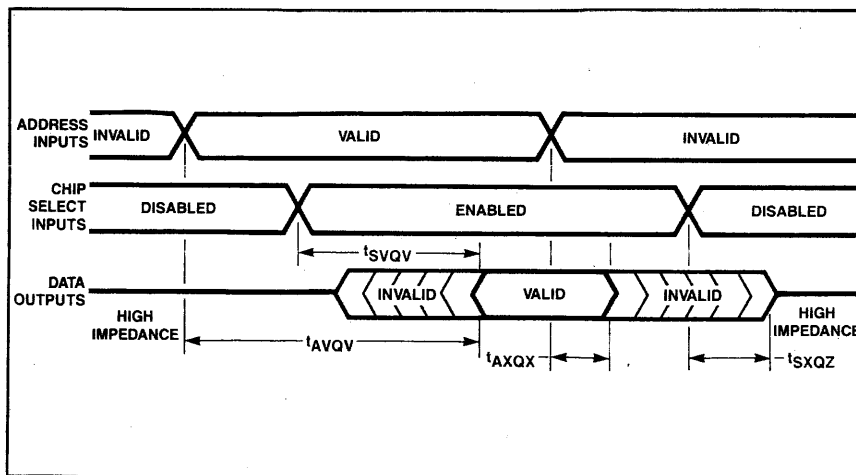
## AC ELECTRICAL CHARACTERISTICS $V_{CC} = 5.0V \pm 10\%$

PARAMETER	26S64-30		26S64-45		UNIT
	Min	Max	Min	Max	
$t_{AVQV}$	Address access time (Note 5)			300	ns
$t_{SVQV}$	Chip select delay (Note 6)			120	ns
$t_{SXQZ}$	Chip deselect delay (Note 7)			100	ns
$t_{AXQX}$	Previous data valid after address change delay		0	0	ns

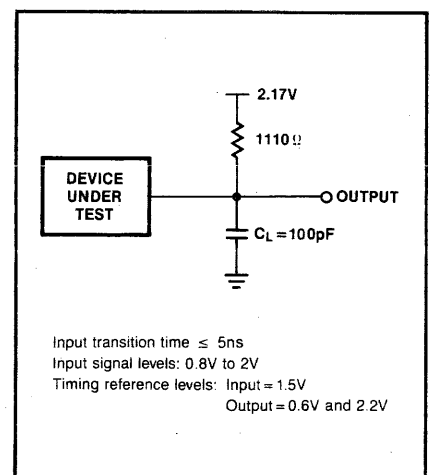
NOTES

- 5. Following a change of banks, at least two times  $t_{AVQV}$  must elapse before a valid access may be exercised.
- 6.  $t_{SVQV}$  is at value indicated, provided the valid address leads the chip select by  $(t_{AVQV} - t_{SVQV})$  ns or more.
- 7.  $t_{SXQZ}$  is measured at  $V_{OL} = 0.5V$  on the "0" to High-Z state transition, with AC test load connected to the output.
- 8. Bank select can be effected only while chip is selected, and bank switch address is held valid for at least 100ns.
- 9. At power-up the 26S64 may come up in either bank; proper initialization is required.

### TIMING DIAGRAM



### AC TEST LOAD



# Si Rotating Memory Integrated Circuit Capabilities

Silicon Systems is a semiconductor company dedicated primarily to the design and production of "application specific" integrated circuits. SSI sometimes works closely with a group of equipment manufacturers to develop an IC. Often these IC's become the "standard products" of the industry. Such is the case with Silicon Systems' family of read/write IC's for the rotating memory market.

## SSI's first read/write Winchester IC developed in 1976

In 1976 Silicon Systems developed its first read/write monolithic bipolar integrated circuit for the IBM 3350 Winchester disk drive PCM (plug-compatible memory) market. Called the SSI 104, it integrated all of the required read, write, control, and data protection functions on one chip. Assembled in a flat pack, the device could be mounted directly on the head-arm assembly. The SSI 104 soon became the industry standard.

## A broad line of Rotating Memory IC's

SSI now has several varieties of read/write products

including the SSI 115 with 2, 4, or 5 channels, the SSI 117 with 4 to 6 channels, and the SSI 501 with 8 channels. For thin-film-head applications, SSI has the 4-channel SSI 114. SSI also has servo preamps including the SSI 101 for ferrite-head drives and the SSI 116 for thin-film-head drives. SSI has defined a floppy-disk read/write IC, the soon-to-be released SSI 570.

## IC's for streamers, tape drives, and controllers, too

If you don't see the product you need, it may be on our standard product list soon. We serve the disk drive market more completely than anyone else with our standard line of circuits, and we have also developed a score of custom IC's of various designs and functions for the rotating memory market: IC's for spindle control, data paths, IC's for data separation and controllers too. So, if the exact circuit you need isn't already in our line — we have the technology and experience to make a custom chip for you.

## SSI PRODUCT SELECTOR GUIDE DISK DRIVE CIRCUITS

Part No.	Circuit Function	No. of Channels	Power Supplies	Logic Type	Data Write/Read	Write Current Source	Write Unsafe Indicator	Write Read Gain	Special Feature
<b>FERRITE</b>									
SSI 101	Servo Preamp	N/A	8.3V 10V	—	—	—	—	77 to 110	Min BW 10 MHz
SSI 104	Read/Write Amp	4	+6V -4V	ECL	Bi-Directional Differential	External	Low for Write Unsafe	28 to 43	
SSI104L	Read/Write Amp	4	+6V -4V	ECL	Bi-Directional Differential	External	Low for Write Unsafe	28 to 43	Low Noise (1.7nV/√Hz)
SSI 108	Read/Write Amp	4	+6V -4V	ECL	Bi-Directional Differential	External	Low for Write Unsafe	28 to 43	24 pin DIP opt 104
SSI 115	Read/Write Amp	2,4,5	±5V	TTL	Bi-Directional Differential	External	Low for Write Unsafe	26 to 52	Write Current Diverter in Read Mode
SSI 117	Read/Write Amp	4,6	+5V +12V	TTL	TTL Write Diff. Read	On-Chip	Low for Write Safe	80 to 120	"OEM" supplies
SSI 122	Read/Write Amp	4	+6V -4V	ECL	Bi-Directional Differential	External	Low for Write Unsafe	28 to 43	22 pin DIP opt 104
SSI 501	Read/Write Amp	8	+5V +12V	TTL	TTL Write Diff. Read	On-Chip	Low for Write Safe	80 to 120	"OEM" supplies
<b>THIN-FILM</b>									
SSI 114	Read/Write Amp	4	±5V	TTL	Diff. Write Diff. Read	On-Chip	Low for Write Safe	75 to 120	Sep. Read/Write Bus, Write Safe Verify output
SSI 116	Servo Preamp	N/A	8.3V	—	—	—	—	200 to 310	Min BW 20 MHz
<b>TAPE DRIVE</b>									
SSI 8500	CRC Character Generator	—	—	—	—	—	—	—	—
SSI 8502	LRC Character Generator	—	—	—	—	—	—	—	—
SSI 8520	Queue/Deskew Register	—	—	—	—	—	—	—	—
<b>FLOPPY READ/WRITE</b>									
SSI 570	Read/Write System	2	+5V +12V	TTL	—	On-Chip	—	—	—

Note: On-Chip Write Current Sources are programmed with an External Resistor.



14351 Myford Road, Tustin, California 92680 (714) 731-7110, TWX 910-595-2809

# ABBREVIATIONS OF COMPANY NAMES

<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.			<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Harris</b>	Harris Semiconductor	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Heurikon</b>	Heurikon Corp.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Hitachi</b>	Hitachi America, Ltd.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>Holt</b>	Holt Inc.		
<b>Analogic</b>	Analogic	<b>HP</b>	Hewlett-Packard		
<b>Analog Sys</b>	Analog Systems	<b>Hughes</b>	Hughes Aircraft, Solid State Products	<b>Panasonic</b>	Panasonic
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Pico Design</b>	Pico Design
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>Polycore</b>	Polycore Electronics
<b>APM</b>	Applied Microsystems Corp.			<b>Plessey</b>	Plessey Semiconductors
<b>Appl Sys</b>	Applied Systems Corp.			<b>PMI</b>	Precision Monolithics, Inc.
<b>APT</b>	Applied Microtechnology	<b>ICC</b>	International Cybernetics	<b>PragDes</b>	Pragmatic Design Inc.
<b>Aptek</b>	Aptek Microsystems	<b>IDT</b>	Integrated Device Technology	<b>Pro-Log</b>	Pro-Log Corp.
<b>Array Tech</b>	Array Technology	<b>IMI</b>	International Microcircuits, Inc.		
<b>AWI</b>	AWI Electronics	<b>IMP</b>	International Microelectronic Products	<b>Quay</b>	Quay Corp.
		<b>IMS</b>	Industrial MicroSystems Inc.		
<b>Barvon</b>	Barvon Research	<b>Infosphere</b>	Infosphere	<b>Raytheon</b>	Raytheon Semiconductor
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>Inmos</b>	Inmos	<b>RCA</b>	RCA Solid State Division
<b>Burr-Brown</b>	Burr-Brown	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>RCI Data</b>	RCI Data
		<b>IntCirSys</b>	Integrated Circuit Systems	<b>RELMS</b>	Relational Memory Systems
<b>CAE</b>	Computer Aided Engineering	<b>IntCompSys</b>	Integrated Computer Systems	<b>Reticon</b>	Reticon
<b>Cal Devices</b>	California Devices	<b>Int Tech</b>	Integrated Technology Corp.	<b>RIFA</b>	RIFA
<b>Cermetek</b>	Cermetek	<b>Intech</b>	Intech Microcircuits Div.	<b>Rockwell</b>	Rockwell, Microelectronic Devices
<b>CGRS</b>	CGRS Microtech Inc.	<b>Intel</b>	Intel	<b>RTC</b>	Riehl Time Corporation
<b>Cherry</b>	Cherry Semiconductor	<b>Interdesign</b>	Interdesign		
<b>CIC</b>	Custom Integrated Circuits	<b>Intersil</b>	Intersil	<b>Sanyo</b>	Sanyo
<b>CirTech</b>	Circuit Technology	<b>Intronics</b>	Intronics	<b>SBE</b>	SBE, Inc.
<b>Citel</b>	Citel	<b>ITT</b>	ITT Semiconductors	<b>SEEQ</b>	SEEQ Technology, Inc.
<b>Comlinear</b>	Comlinear Corporation			<b>SPI</b>	Semi Processes Inc.
<b>CMA</b>	Custom MOS Arrays	<b>Kinetic Sys</b>	Kinetic Systems	<b>Siemens</b>	Siemens
<b>Comark</b>	Comark Corp.	<b>Kontron</b>	Kontron Electronics	<b>Si-Fab</b>	Si-Fab
<b>Comdial</b>	Comdial Semiconductor			<b>Signetics</b>	Signetics
<b>Comp Auto</b>	Computer Automation	<b>Lambda</b>	Lambda Semiconductor	<b>SGS</b>	SGS Semiconductor
<b>Compas</b>	Compas Microsystems	<b>Linear Tech</b>	Linear Technology	<b>Sharp</b>	Sharp
<b>Cont Logic</b>	Control Logic Inc.	<b>LSI Comp</b>	LSI Computer Systems	<b>Silicon G</b>	Silicon General
<b>Control Sys</b>	Control Systems Microsystems Div.	<b>LSI Logic</b>	LSI Logic Corporation	<b>Siliconix</b>	Siliconix
<b>CreMicro</b>	Creative Micro Systems			<b>Silicon Sys</b>	Silicon Systems Inc.
<b>Cromemco</b>	Cromemco, Inc.	<b>Master Logic</b>	Master Logic Corporation	<b>Siltronics</b>	Siltronics
<b>Cubit</b>	Cubit Inc.	<b>Matrix</b>	Matrix Corp.	<b>Standard Microsystems Corp.</b>	Standard Microsystems Corp.
<b>Curtis</b>	Curtis Electro Devices, Inc.	<b>Matrox</b>	Matrox Electronic Systems	<b>S MOS</b>	S MOS Systems
<b>Cybernetic</b>	Cybernetic Micro Systems	<b>Maxim</b>	Maxim Integrated Products	<b>Solarise</b>	Solarise Enterprises
<b>Cybersys</b>	Cybersystems	<b>MCC</b>	Micro-Computer Control	<b>Solitron</b>	Solitron Devices
<b>Cybertek</b>	Cybertek Inc.	<b>MCE</b>	MCE Electronics	<b>Sprague</b>	Sprague Electric Company
		<b>Micrel</b>	Micrel	<b>SSM</b>	Solid State Micro Technology for Music
<b>Data General</b>	Data General	<b>Micro Innov</b>	Micro Innovators	<b>SSS</b>	Solid State Scientific
<b>Data I/O</b>	Data I/O	<b>Micropac</b>	Micropac Industries	<b>Stag</b>	Stag Microsystems
<b>Data Trans</b>	Data Translation	<b>Micro Net</b>	Micro Networks	<b>STC</b>	Storage Technology Corp.
<b>Datel</b>	Datel	<b>Micro Pwr</b>	Micro Power Systems	<b>STD</b>	STD Microsystems
<b>Datronic</b>	Datronic Corporation	<b>Micro Sci</b>	Micro Sciences Corp.	<b>Struc Des</b>	Structured Design Inc.
<b>DDC</b>	Data Devices Corporation	<b>Micro Tech</b>	Microcircuits Technology	<b>Stynetic</b>	Stynetic Systems
<b>DEC</b>	Digital Equipment Corporation	<b>Micro-Link</b>	Micro-Link Corporation	<b>Sunrise</b>	Sunrise Electronics
<b>Die-Tech</b>	Die-Tech	<b>Micron</b>	Micron Technology	<b>Sunshine</b>	Sunshine Semiconductor
<b>Digelec</b>	Digelec Corp.	<b>MilerTron</b>	MilerTronics	<b>Supertex</b>	Supertex Inc.
<b>Digitek</b>	Digitek, Inc.	<b>Miller</b>	Miller Technology	<b>Symtek</b>	Symtek Corp.
<b>Dionics</b>	Dionics Inc.	<b>Mitel</b>	Mitel Semiconductor	<b>Synertek</b>	Synertek
<b>Dist Comp</b>	Distributed Computer Systems	<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Sys Innov</b>	Systems Innovations
<b>Divers Tech</b>	Diversified Technology	<b>MMI</b>	Monolithic Memories, Inc.		
		<b>Mostek</b>	Monolithic Systems Corp.	<b>Tau Zero</b>	Tau Zero Inc.
<b>E-HI</b>	E-H International, Inc.	<b>Motorola</b>	Motorola Semiconductor	<b>Technitrol</b>	Technitrol
<b>EDI</b>	Electronic Designs Inc.	<b>MRC</b>	MRC Systems	<b>Tektronix</b>	Tektronix
<b>Elind</b>	Elind Electronica Industriale	<b>Murray</b>	Murray Consulting	<b>Teledyne C</b>	Teledyne Crystalonics
<b>EMM-SESCO</b>	EEM-SESCO	<b>Monosil</b>	Monolithic Systems Corp.	<b>Teledyne P</b>	Teledyne Philbrick
<b>Emulogic</b>	Emulogic Inc.			<b>Teledyne S</b>	Teledyne Semiconductor
<b>ETI Micro</b>	ETI Micro	<b>National</b>	National Semiconductor	<b>Telefunken</b>	Telefunken
<b>Exar</b>	Exar Integrated Systems	<b>NCM</b>	NCM Corp.	<b>Telmos</b>	Telmos
<b>Exel</b>	Exel Microelectronics	<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Teltone</b>	Teltone Corporation
		<b>NEC</b>	NEC Electronics	<b>TI</b>	Texas Instruments
<b>Fairchild</b>	Fairchild	<b>Nitron</b>	Nitron	<b>Third Domain</b>	Third Domain
<b>Ferranti</b>	Ferranti Electric			<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
<b>Force</b>	Force Computers			<b>Toshiba</b>	Toshiba America
<b>Fujitsu A</b>	Fujitsu America			<b>Trans-Data</b>	Trans-Data
<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.			<b>TRW</b>	TRW LSI Products
				<b>Unitrode</b>	Unitrode
				<b>Universal</b>	Universal Semiconductor, Inc.
				<b>Varix</b>	Varix Corp.
				<b>VLSI Design</b>	VLSI Design Associates
				<b>VTI</b>	VLSI Technology, Inc.
				<b>Votrax</b>	Votrax
				<b>Weitek</b>	Weitek Corporation
				<b>Western</b>	Western Digital
				<b>Wintek</b>	Wintek Corp.
				<b>Xicor</b>	Xicor, Inc.
				<b>Xycom</b>	Xycom
				<b>Zendex</b>	Zendex Corp.
				<b>Zilog</b>	Zilog
				<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrex</b>	Zytrex Corp.

# RAM Selection Guide

Commercial  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$

Part Number	Organization	Access Time (ns)	Maximum Current (mA)		Power Supply (Volts)	Number of Pins	Package Type (Note 1)
			Operating	Standby			
SY2101-1	256 x 4	500	70	--	+5	22	C, P
SY2101A	256 x 4	350	55	--	+5	22	C, P
SY2101A-2	256 x 4	250	55	--	+5	22	C, P
SY2101A-4	256 x 4	450	55	--	+5	22	C, P
SY2111-1	256 x 4	500	70	--	+5	18	C, D, P
SY2111A	256 x 4	350	55	--	+5	18	C, D, P
SY2111A-2	256 x 4	250	55	--	+5	18	C, D, P
SY2111A-4	256 x 4	450	55	--	+5	18	C, D, P
SY2112-1	256 x 4	500	70	--	+5	16	C, D, P
SY2112A	256 x 4	350	55	--	+5	16	C, D, P
SY2112A-2	256 x 4	250	55	--	+5	16	C, D, P
SY2112A-4	256 x 4	450	55	--	+5	16	C, D, P
SY2114AL-1	1024 x 4	100	40	--	+5	18	C, D, P
SY2114AL-2	1024 x 4	120	40	--	+5	18	C, D, P
SY2114AL-3	1024 x 4	150	40	--	+5	18	C, D, P
SY2114AL-4	1024 x 4	200	40	--	+5	18	C, D, P
SY2114A-4	1024 x 4	200	70	--	+5	18	C, D, P
SY2114A-5	1024 x 4	250	70	--	+5	18	C, D, P
SY2148H	1024 x 4	70	150	30	+5	18	C, D, P
SY2148H-2	1024 x 4	45	150	30	+5	18	C, D, P
SY2148H-3	1024 x 4	55	150	30	+5	18	C, D, P
SY2148HL	1024 x 4	70	125	20	+5	18	C, D, P
SY2148HL-3	1024 x 4	55	125	20	+5	18	C, D, P
SY2149H	1024 x 4	70	150	--	+5	18	C, D, P
SY2149H-2	1024 x 4	45	150	--	+5	18	C, D, P
SY2149H-3	1024 x 4	55	150	--	+5	18	C, D, P
SY2149HL	1024 x 4	70	125	--	+5	18	C, D, P
SY2149HL-3	1024 x 4	55	125	--	+5	18	C, D, P
SY2147H-2	4096 x 1	45	180	30	+5	18	C, D, P
SY2147H-3	4096 x 1	55	180	30	+5	18	C, D, P
SY2147H	4096 x 1	70	160	20	+5	18	C, D, P
SY2147HL-3	4096 x 1	55	125	15	+5	18	C, D, P
SY2147HL	4096 x 1	70	140	10	+5	18	C, D, P
SY2158-2	1024 x 8	120	100	30	+5	24	P
SY2158-3	1024 x 8	150	100	30	+5	24	P
SY2158-4	1024 x 8	200	100	30	+5	24	P
SY2159-2	1024 x 8	120	100	--	+5	24	P
SY2159-3	1024 x 8	150	100	--	+5	24	P
SY2159-4	1024 x 8	200	100	--	+5	24	P
SY2128-1	2048 x 8	100	100	20	+5	24	K, C, D, P
SY2128-2	2048 x 8	120	100	20	+5	24	K, C, D, P
SY2128-3	2048 x 8	150	100	20	+5	24	K, C, D, P
SY2128-4	2048 x 8	200	100	20	+5	24	K, C, D, P
SY2128L-1	2048 x 8	100	80	15	+5	24	K, C, D, P
SY2128L-2	2048 x 8	120	80	15	+5	24	K, C, D, P
SY2128L-3	2048 x 8	150	80	15	+5	24	K, C, D, P
SY2128L-4	2048 x 8	200	80	15	+5	24	K, C, D, P
SY2129-1	2048 x 8	100	100	--	+5	24	K, C, D, P
SY2129-2	2048 x 8	120	100	--	+5	24	K, C, D, P
SY2129-3	2048 x 8	150	100	--	+5	24	K, C, D, P
SY2129-4	2048 x 8	200	100	--	+5	24	K, C, D, P
SY2129L-1	2048 x 8	100	80	--	+5	24	K, C, D, P
SY2129L-2	2048 x 8	120	80	--	+5	24	K, C, D, P
SY2129L-3	2048 x 8	150	80	--	+5	24	K, C, D, P
SY2129L-4	2048 x 8	200	80	--	+5	24	K, C, D, P
SY2168 [2]	4096 x 4	70	120	30	+5	20	C, D, P
SY2168-3 [2]	4096 x 4	55	120	30	+5	20	C, D, P
SY2169 [2]	4096 x 4	70	120	--	+5	20	C, D, P
SY2169-3 [2]	4096 x 4	55	120	--	+5	20	C, D, P
SY2167 [2]	16,384 x 1	70	120	30	+5	20	C, D, P
SY2167-3 [2]	16,384 x 1	55	120	30	+5	20	C, D, P
SY2130 [2]	1024 x 8	100	170	40	+5	48	C, D, P
SY2131 [2]	1024 x 8	100	170	--	+5	48	C, P

NOTES: 1. C = Ceramic, D = Cerdip, P = Plastic, K = Leadless Chip Carrier.  
 2. Preliminary Information (new product).

# ROM Selection Guide

Commercial:  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$

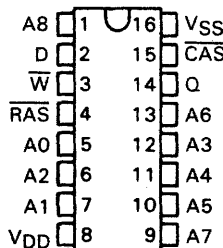
Part Number	Organization	Access Time (ns) Max.	Maximum Current (mA)		Power Supply (Volts)	Number of Pins	Package Type (Note 1)	Compatible EPROM/PROM
			Operating	Standby				
SY3308	1024 x 8	70	120	—	+5	24	C, D, P, K	82S181
SY3308R	1024 x 8	35 <sup>[3]</sup>	130	—	+5	24	C, D, P, K	27S35
SY2316B	2048 x 8	450	98	—	+5	24	C, D, P	2716
SY2316B-2	2048 x 8	200	98	—	+5	24	C, D, P	2716
SY2316B-3	2048 x 8	300	98	—	+5	24	C, D, P	2716
SY3316	2048 x 8	80	120	—	+5	24	C, D, P, K	82S191
SY3316A	2048 x 8	80	120	20	+5	24	C, D, P, K	82S191
SY3316R	2048 x 8	35 <sup>[3]</sup>	130	—	+5	24	C, D, P, K	27S45
SY2332	4096 x 8	450	100	—	+5	24	C, D, P	TMS2532
SY2332-2	4096 x 8	200	100	—	+5	24	C, D, P	TMS2532
SY2332-3	4096 x 8	300	100	—	+5	24	C, D, P	TMS2532
SY2333	4096 x 8	450	100	—	+5	24	C, D, P	2732/A
SY2333-2	4096 x 8	200	100	—	+5	24	C, D, P	2732/A
SY2333-3	4096 x 8	300	100	—	+5	24	C, D, P	2732/A
SY2364	8192 x 8	450	100	—	+5	24	C, D, P	TMS2564
SY2364-2	8192 x 8	200	100	—	+5	24	C, D, P	TMS2564
SY2364-3	8192 x 8	300	100	—	+5	24	C, D, P	TMS2564
SY2364A	8192 x 8	450	100	12	+5	24	C, D, P	TMS2564
SY2364A-2	8192 x 8	200	100	12	+5	24	C, D, P	TMS2564
SY2364A-3	8192 x 8	300	100	12	+5	24	C, D, P	TMS2564
SY2365	8192 x 8	450	100	—	+5	28	C, D, P	2764
SY2365-2	8192 x 8	200	100	—	+5	28	C, D, P	2764
SY2365-3	8192 x 8	300	100	—	+5	28	C, D, P	2764
SY2365A	8192 x 8	450	100	12	+5	28	C, D, P	2764
SY2365A-2	8192 x 8	200	100	12	+5	28	C, D, P	2764
SY2365A-3	8192 x 8	300	100	12	+5	28	C, D, P	2764
SY23128-2	16,384 x 8	200	100	—	+5	28	C, D, P	27128
SY23128-3	16,384 x 8	300	100	—	+5	28	C, D, P	27128
SY23128	16,384 x 8	450	100	—	+5	28	C, D, P	27128
SY23128A-2	16,384 x 8	200	100	10	+5	28	C, D, P	27128
SY23128A-3	16,384 x 8	300	100	10	+5	28	C, D, P	27128
SY23128A	16,384 x 8	450	100	10	+5	28	C, D, P	27128
SY23130	16,384 x 8	450	100	—	+5	28	C, D, P	—
SY23130-2	16,384 x 8	200	100	—	+5	28	C, D, P	—
SY23130-3	16,384 x 8	300	100	—	+5	28	C, D, P	—
SY23130A	16,384 x 8	450	100	10	+5	28	C, D, P	—
SY23130A-2	16,384 x 8	200	100	10	+5	28	C, D, P	—
SY23130A-3	16,384 x 8	300	100	10	+5	28	C, D, P	—
SY23256-2	32,768 x 8	200	100	—	+5	28	C, D, P	27256
SY23256-3	32,768 x 8	300	100	—	+5	28	C, D, P	27256
SY23256	32,768 x 8	450	100	—	+5	28	C, D, P	27256
SY23256A-2	32,768 x 8	200	100	10	+5	28	C, D, P	27256
SY23256A-3	32,768 x 8	300	100	10	+5	28	C, D, P	27256
SY23256A	32,768 x 8	450	100	10	+5	28	C, D, P	27256
SY6364 <sup>[2]</sup>	8192 x 8	200	70	1mA/10 $\mu$ A <sup>[4]</sup>	+5	24	C, D, P	
SY6365 <sup>[2]</sup>	8192 x 8	200	70	1mA/10 $\mu$ A <sup>[4]</sup>	+5	28	C, D, P	

**NOTES:**

1. C = Ceramic, D = Cerdip, P = Plastic, K = Leadless Chip Carrier.
2. Preliminary information.
3. Effective Access Time ( $t_{CPA}$ ).
4.  $\overline{CE}$  @ 2V/ $\overline{CE}$  @  $V_{CC}$ .

- 262,144 X 1 Organization
- Single +5-V Supply (10% Tolerance)
- JEDEC Standardized Pin Out
- Upward Pin Compatible with TMS4164 (64K Dynamic RAM)
- Performance Ranges:

TMS4256, TMS4257 . . . JL OR NL PACKAGE  
(TOP VIEW)



DEVICE	ACCESS TIME ROW ADDRESS (MAX)	ACCESS TIME COLUMN ADDRESS (MAX)	READ OR WRITE CYCLE (MIN)
TMS4256-10	100 ns	50 ns	200 ns
TMS4257-10			
TMS4256-12	120 ns	60 ns	230 ns
TMS4257-12			
TMS4256-15	150 ns	75 ns	260 ns
TMS4257-15			
TMS4256-20	200 ns	100 ns	330 ns
TMS4257-20			

PIN NOMENCLATURE	
A0-A8	Address Inputs
$\overline{CAS}$	Column Address Strobe
D	Data-In
Q	Data-Out
$\overline{RAS}$	Row Address Strobe
$\overline{W}$	Write Enable
VDD	+5-V Supply
VSS	Ground

- Long Refresh Period . . . 4 ms (MAX)
- Low Refresh Overhead Time . . . As Low As 1.3% of Total Refresh Period
- On-Chip Substrate Bias Generator
- All Inputs, Outputs, and Clocks Fully TTL Compatible
- 3-State Unlatched Output
- Common I/O Capability with "Early Write" Feature
- Page ('4256) or Nibble-Mode ('4257) Options for Faster Access Operation
- Low Power Dissipation:  
Operating . . . 200 mW (TYP)  
Standby . . . 16.5 mW (TYP)
- $\overline{RAS}$  Only Refresh Mode
- Hidden Refresh Mode

description

The '4256 and '4257 are high-speed, 262,144-bit dynamic random-access memories, organized as 262,144 words of one bit each. They employ state-of-the-art SMOS (scaled MOS) N-channel double-level polysilicon gate technology for very high performance combined with low cost and improved reliability.

These devices feature maximum  $\overline{RAS}$  access times of 100 ns, 120 ns, 150 ns, or 200 ns. Typical power dissipation is 200 mW operating and 16.5 mW standby.

New SMOS technology permits operation from a single +5-V supply, reducing system power supply and decoupling requirements, and easing board layout.  $I_{DD}$  peaks are 150 mA typical, and a -1-V input voltage undershoot can be tolerated, minimizing system noise considerations.

All inputs and outputs, including clocks, are compatible with Series 74 TTL. All address and data-in lines are latched on chip to simplify system design. Data-out is unlatched to allow greater system flexibility.

PRODUCT PREVIEW

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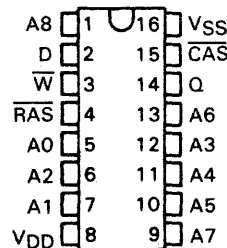


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- 262,144 X 1 Organization
- Single +5-V Supply (10% Tolerance)
- JEDEC Standardized Pin Out
- Upward Pin Compatible with TMS4164 (64K Dynamic RAM)
- Performance Ranges:

TMS4256, TMS4257 . . . JL OR NL PACKAGE  
(TOP VIEW)



DEVICE	ACCESS TIME ROW ADDRESS (MAX)	ACCESS TIME COLUMN ADDRESS (MAX)	READ OR WRITE CYCLE (MIN)
TMS4256-10 TMS4257-10	100 ns	50 ns	200 ns
TMS4256-12 TMS4257-12	120 ns	60 ns	230 ns
TMS4256-15 TMS4257-15	150 ns	75 ns	260 ns
TMS4256-20 TMS4257-20	200 ns	100 ns	330 ns

PIN NOMENCLATURE	
A0-A8	Address Inputs
$\overline{\text{CAS}}$	Column Address Strobe
D	Data-In
Q	Data-Out
$\overline{\text{RAS}}$	Row Address Strobe
$\overline{\text{W}}$	Write Enable
VDD	+5-V Supply
VSS	Ground

- Long Refresh Period . . . 4 ms (MAX)
- Low Refresh Overhead Time . . . As Low As 1.3% of Total Refresh Period
- On-Chip Substrate Bias Generator
- All Inputs, Outputs, and Clocks Fully TTL Compatible
- 3-State Unlatched Output
- Common I/O Capability with "Early Write" Feature
- Page ('4256) or Nibble-Mode ('4257) Options for Faster Access Operation
- Low Power Dissipation:  
Operating . . . 200 mW (TYP)  
Standby . . . 16.5 mW (TYP)
- $\overline{\text{RAS}}$  Only Refresh Mode
- Hidden Refresh Mode

description

The '4256 and '4257 are high-speed, 262,144-bit dynamic random-access memories, organized as 262,144 words of one bit each. They employ state-of-the-art SMOS (scaled MOS) N-channel double-level polysilicon gate technology for very high performance combined with low cost and improved reliability.

These devices feature maximum  $\overline{\text{RAS}}$  access times of 100 ns, 120 ns, 150 ns, or 200 ns. Typical power dissipation is 200 mW operating and 16.5 mW standby.

New SMOS technology permits operation from a single +5-V supply, reducing system power supply and decoupling requirements, and easing board layout.  $I_{DD}$  peaks are 150 mA typical, and a -1-V input voltage undershoot can be tolerated, minimizing system noise considerations.

All inputs and outputs, including clocks, are compatible with Series 74 TTL. All address and data-in lines are latched on chip to simplify system design. Data-out is unlatched to allow greater system flexibility.

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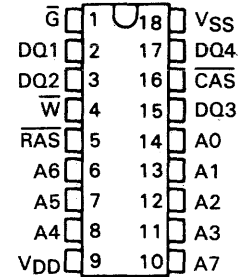
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- 65,536 X 4 Organization
- Single +5-V Supply (10% Tolerance)
- JEDEC Standardized Pin-Out
- Pin-Out Identical to TMS4416 (16K X 4 Dynamic RAM)
- Performance Ranges:

DEVICE	ACCESS TIME ROW ADDRESS (MAX)	ACCESS TIME COLUMN ADDRESS (MAX)	READ OR WRITE CYCLE (MIN)	READ-MODIFY-WRITE CYCLE (MIN)
TMS4464-10	100 ns	60 ns	200 ns	275 ns
TMS4464-12	120 ns	70 ns	230 ns	320 ns
TMS4464-15	150 ns	85 ns	260 ns	365 ns
TMS4464-20	200 ns	120 ns	330 ns	460 ns

- Long Refresh Period . . . 4 ms (MAX)
- Low Refresh Overhead Time . . . As Low As 1.3% of Total Refresh Period
- On-Chip Substrate Bias Generator
- All Inputs, Outputs, and Clocks Fully TTL Compatible
- 3-State Unlatched Output
- Early Write or  $\bar{G}$  to Control Output Buffer Impedance
- Page-Mode Operation for Faster Access
- Low Power Dissipation:
  - Operating . . . 350 mW (TYP)
  - Standby . . . 16.5 mW (TYP)
- $\overline{RAS}$  Only Refresh Mode
- Hidden Refresh Mode

TMS4464 . . . JL OR NL PACKAGE (TOP VIEW)



PIN NOMENCLATURE	
A0-A7	Address Inputs
CAS	Column Address Strobe
DQ1-DQ4	Data-In/Data-Out
$\bar{G}$	Output Enable
$\overline{RAS}$	Row Address Strobe
VDD	+5-V Supply
VSS	Ground
$\bar{W}$	Write Enable

**description**

The TMS4464 is a high-speed, 262,144-bit dynamic random-access memory, organized as 65,536 words of four bits each. It employs state-of-the-art SMOS (scaled MOS) N-channel double-level polysilicon gate technology for very high performance combined with low cost and improved reliability.

This device features maximum  $\overline{RAS}$  access times of 100 ns, 120 ns, 150 ns, or 200 ns. Typical power dissipation is 350 mW operating and 16.5 mW standby.

New SMOS technology permits operation from a single +5-V supply, reducing system power supply and decoupling requirements, and easing board layout.  $I_{DD}$  peaks are 150 mA typical, and a -1-V input voltage undershoot can be tolerated, minimizing system noise considerations.

All inputs and outputs, including clocks, are compatible with Series 74 TTL. All address and data-in lines are latched on chip to simplify system design. Data-out is unlatched to allow greater system flexibility.

The TMS4464 is offered in an 18-pin dual-in-line ceramic or plastic package and is guaranteed for operation from 0°C to 70°C. These packages are designed for insertion in mounting-hole rows on 300-mil (7.62 mm) centers.

**PRODUCT PREVIEW**

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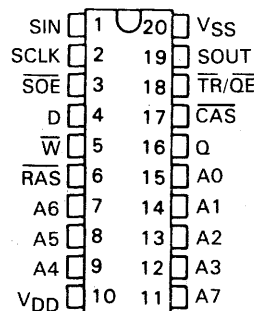
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**TEXAS  
INSTRUMENTS**

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- Dual Accessibility – One Port Sequential Access, One Port Random Access
- Four Cascaded 64-Bit Serial Shift Registers for Sequential Access Applications
- Shift Register Loaded Once Every 64, 128, 192, or 256 Shift Cycles as Desired by User
- Fast Serial Port . . . 25 MHz Shift Rate
- $\overline{\text{TR/QE}}$  as Output Enable Allows Direct Connection of D, Q and Address Lines to Simplify System Design
- Random Access Port Looks Exactly Like a TMS4164
- Separate Serial In and Serial Out to Allow Simultaneous Shift In and Out
- 65,536 × 1 Organization
- Maximum Access Time from  $\overline{\text{RAS}}$  Less Than 150 ns
- Minimum Cycle Time (Read or Write) Less Than 260 ns
- Long Refresh Period . . . 4 Milliseconds
- Low Refresh Overhead Time . . . As Low As 1.6% of Total Refresh Period
- All Inputs, Outputs, Clocks Fully TTL Compatible
- 3-State Unlatched Outputs for Both Random and Serial Access
- Common I/O Capability with "Early Write" Feature
- Page-Mode Operation for Faster Access
- Low Power Dissipation
  - Operating . . . 200 mW (Typical)
  - Standby . . . 35 mW (Typical)
- New SMOS (Scaled-MOS) N-Channel Technology
- $\overline{\text{SOE}}$  Simplifies Multiplexing of Video Data Streams

TMS4161 . . . NL PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A0-A7	Address Inputs
$\overline{\text{CAS}}$	Column Address Strobe
D	Random Access Data-In
Q	Random Access Data-Out
$\overline{\text{RAS}}$	Row Address Strobe
SCLK	Serial Data Clock
SIN	Serial Data-In
$\overline{\text{SOE}}$	Serial Output Enable
SOUT	Serial Data-Out
$\overline{\text{TR/QE}}$	Register Transfer/Q Output Enable
W	Write Enable
VDD	+5-V Supply
VSS	Ground

description

The TMS4161 is a high-speed, dual-access 65,536-bit dynamic random-access memory. The random-access port makes the memory look like it is organized as 65,536 words of one bit each like the TMS4164. The sequential access port is interfaced to an internal 256-bit dynamic shift register organized as four 64-bit shift registers which makes the memory look like it is organized as up to 256 words of up to 256 bits each which are accessed serially. One,

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TEXAS  
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## TMS4161 65,536-BIT MULTIPOINT MEMORY

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two, three, or four 64-bit shift registers can be sequentially read out depending on a two-bit code applied to the two most significant column address inputs. The TMS4161 employs state-of-the-art SMOS (Scaled-MOS) N-channel double level polysilicon gate technology for very high performance combined with low cost and improved reliability.

The TMS4161 features full asynchronous dual access capability except when transferring data between the shift register and the memory array.

Refresh period is extended to 4 milliseconds, and during this period each of the 256 rows must be strobed with  $\overline{\text{RAS}}$  in order to retain data.  $\overline{\text{CAS}}$  can remain high during the refresh sequence to conserve power. Note that the transfer of a row of data from the memory array to the shift register also refreshes that row.

All inputs and outputs, including clocks, are compatible with Series 74 TTL. All address lines and data-in are latched on chip to simplify system design. Data-out is unlatched to allow greater system flexibility.

The TMS4161 is offered in a 20-pin dual-in-line-plastic package and is guaranteed for operation from 0°C to 70°C. Packages are designed for insertion in mounting-hole rows on 300-mil (7.62 mm) centers.

### random access address space to sequential address space mapping

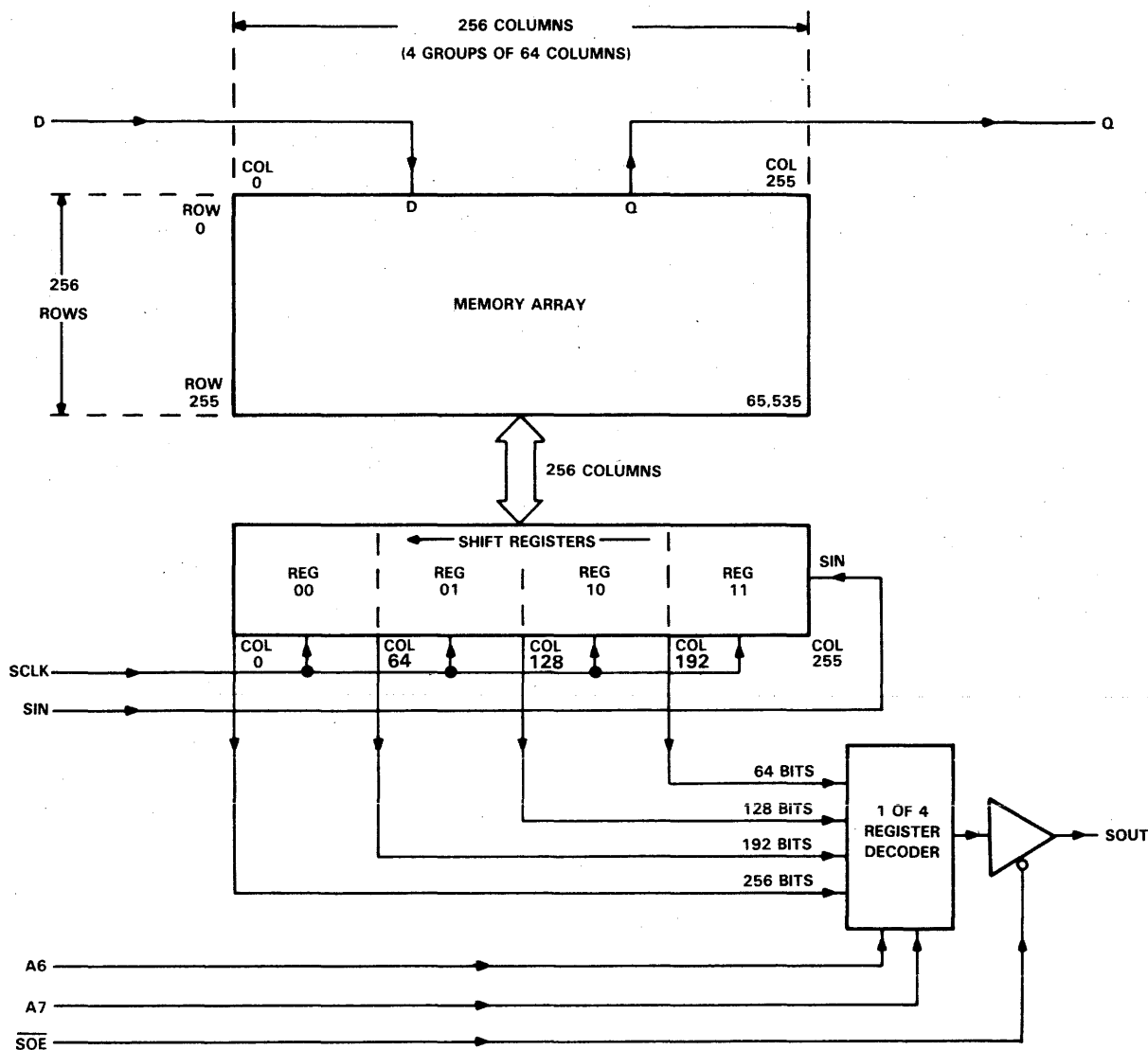
The TMS4161 is designed with each row divided into four, 64-column sections. The first column section to be shifted out is selected by the two most significant column address bits. If the two bits represent binary 00, then one to four registers can be shifted out in order. If the two bits represent binary 01, then only 1 to 3 (the most significant) registers can be shifted out in order. If the two bits represent 10, then one to two of the most significant registers can be shifted out in order. Finally, if the two bits represent 11 only the most significant register can be shifted out. All registers are shifted out with the least significant bit (bit 0) first and the most significant bit (bit 63) last. Note that if the two column address bits equal 00 during the last register transfer cycle ( $\overline{\text{TR}}/\overline{\text{QE}}$  equal to 0) a total of 256 bits can be sequentially read out.

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TEXAS  
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functional block diagram



random access operation

$\overline{TR}/\overline{QE}$

The  $\overline{TR}/\overline{QE}$  pin has two functions. First, it selects either register transfer or random-access operation as  $\overline{RAS}$  falls, and second, if this is a random-access operation, it functions as an output enable after  $\overline{CAS}$  falls.

To use the TMS4161 in the random-access mode,  $\overline{TR}/\overline{QE}$  must be high as  $\overline{RAS}$  falls. Holding  $\overline{TR}/\overline{QE}$  high disconnects the 256 elements of the shift registers from the corresponding 256 bit lines of the memory array. If data is to be shifted, the shift registers must be disconnected from the bit lines. Holding  $\overline{TR}/\overline{QE}$  low enables the 256 switches that connect the shift registers to the bit lines and indicates that a transfer will occur between the shift registers and one of the memory rows.

Once  $\overline{CAS}$  has been pulled low,  $\overline{TR}/\overline{QE}$  controls when the data will appear at the Q output (if this is a read cycle). Whenever  $\overline{TR}/\overline{QE}$  is held high, the Q output will be in the high-impedance state. This feature removes the possibility

# TMS4161

## 65,536-BIT MULTIPORT MEMORY

of an overlap between data on the address lines and data appearing on the Q output making it possible to connect the address lines to the Q and D lines (Use of this organization prohibits the use of the early write cycle.).

### address (A0 through A7)

Sixteen address bits are required to decode 1 of 65,536 storage cell locations. Eight row-address bits are set up on pins A0 through A7 and latched onto the chip by the row-address strobe ( $\overline{\text{RAS}}$ ). Then the eight column-address bits are set up on pins A0 through A7 and latched onto the chip by the column-address strobe ( $\overline{\text{CAS}}$ ). All addresses must be stable on or before the falling edges of  $\overline{\text{RAS}}$  and  $\overline{\text{CAS}}$ .  $\overline{\text{RAS}}$  is similar to a chip enable in that it activates the sense amplifiers as well as the row decoder.  $\overline{\text{CAS}}$  is used as a chip select activating the column decoder and the input and output buffers.

### write enable ( $\overline{\text{W}}$ )

The read or write mode is selected through the write enable ( $\overline{\text{W}}$ ) input. A logic high on the  $\overline{\text{W}}$  input selects the read mode and a logic low selects the write mode. The write enable terminal can be driven from standard TTL circuits without a pull-up resistor. The data input is disabled when the read mode is selected. When  $\overline{\text{W}}$  goes low prior to  $\overline{\text{CAS}}$ , data-out will remain in the high-impedance state for the entire cycle permitting common I/O operation.

### data-in (D)

Data is written during a write or read-modify-write cycle. The falling edge of  $\overline{\text{CAS}}$  or  $\overline{\text{W}}$  strobes data into the on-chip data latch. This latch can be driven from standard TTL circuits without a pull-up resistor. In an early write cycle,  $\overline{\text{W}}$  is brought low prior to  $\overline{\text{CAS}}$  and the data is strobed in by  $\overline{\text{CAS}}$  with setup and hold times referenced to this signal. In a delayed write or read-modify-write cycle,  $\overline{\text{CAS}}$  will already be low, thus the data will be strobed in by  $\overline{\text{W}}$  with setup and hold times referenced to this signal.

### data-out (Q)

The three-state output buffer provides direct TTL compatibility (no pull-up resistor required) with a fan-out of two Series 74 TTL loads. Data-out is the same polarity as data-in. The output is in the high-impedance (floating) state as long as  $\overline{\text{CAS}}$  or  $\overline{\text{TR/QE}}$  is held high. Data will not appear on the output until after both  $\overline{\text{CAS}}$  and  $\overline{\text{TR/QE}}$  have been brought low. In a read cycle, the guaranteed maximum output enable access time is valid only if  $t_{\text{CQE}}$  is greater than  $t_{\text{CQE MAX}}$ , and  $t_{\text{RLCL}}$  is greater than  $t_{\text{RLCL MAX}}$ . Likewise,  $t_{\text{a(C) MAX}}$  is valid only if  $t_{\text{RLCL}}$  is greater than  $t_{\text{RLCL MAX}}$ . Once the output is valid, it will remain valid while  $\overline{\text{CAS}}$  and  $\overline{\text{TR/QE}}$  are both low;  $\overline{\text{CAS}}$  or  $\overline{\text{TR/QE}}$  going high will return the output to a high-impedance state. In an early write cycle, the output is always in a high-impedance state. In a delayed write or read-modify-write cycle, the output will follow the sequence for the read cycle. In a register transfer cycle, the output will always be in a high-impedance state.

### refresh

A refresh operation must be performed at least every four milliseconds to retain data. Since the output buffer is in high-impedance state unless  $\overline{\text{CAS}}$  is applied, the  $\overline{\text{RAS}}$  only refresh sequence avoids any output during refresh. Strobing each of the 256 row addresses (A0 through A7) with  $\overline{\text{RAS}}$  causes all bits in each row to be refreshed.  $\overline{\text{CAS}}$  can remain high (inactive) for this refresh sequence to conserve power. Note that the shift registers are also dynamic storage elements and that the data held in the registers will be lost unless SCLK goes high to shift the data one bit position or else the data is reloaded from the memory array. See specifications for maximum register data retention times.

### page-mode

Page-mode operation allows effectively faster memory access by keeping the same row address and strobing successive column addresses onto the chip. Thus, the time required to setup and strobe sequential row addresses for the same page is eliminated. To extend beyond the 256 column locations on a single RAM, the row address and  $\overline{\text{RAS}}$  are applied to multiple 64K RAMs.  $\overline{\text{CAS}}$  is then decoded to select the proper RAM.

TEXAS  
INSTRUMENTS

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**sequential access operation**

**$\overline{\text{TR/QE}}$**

Memory operations involving parallel use of the shift register are first indicated by bringing  $\overline{\text{TR/QE}}$  low before  $\overline{\text{RAS}}$  falls low. This enables the switches connecting the 256 elements of the shift register to the 256 bit lines of the memory array. The  $\overline{\text{W}}$  line determines whether the data will be transferred from or to the shift registers.

**write enable ( $\overline{\text{W}}$ )**

In the sequential access mode,  $\overline{\text{W}}$  determines whether a transfer will occur from the shift registers to the memory array, or from the memory array to the shift registers. To transfer from the shift registers to the memory array,  $\overline{\text{W}}$  is held low as  $\overline{\text{RAS}}$  falls, and, to transfer from the memory array to the shift registers,  $\overline{\text{W}}$  is held high as  $\overline{\text{RAS}}$  falls. Thus, reads and writes are always with respect to the memory array. The write setup and hold times are referenced to the falling edge of  $\overline{\text{RAS}}$  for this mode of operation.

**row address (A0 through A7)**

Eight address bits are required to select one of the 256 possible rows involved in the transfer of data to or from the shift registers. The A0-A7,  $\overline{\text{W}}$ , and the  $\overline{\text{TR/QE}}$  line are latched on the falling edge of  $\overline{\text{RAS}}$ .

**register column address (A7, A6)**

To select one of the four shift registers (transfer from memory to register only), the appropriate 2-bit column address (A7, A6) must be valid when  $\overline{\text{CAS}}$  falls. However, the  $\overline{\text{CAS}}$  and register address signals need not be supplied every cycle, only when it is desired to change or select a new register.

**SCLK**

Data is shifted in and out on the rising edge of SCLK. This makes it possible to view the shift registers as though it were made of 256 rising edge D flip-flops connected D to Q. The TMS4161 is designed to work with a wide range duty cycle clock to simplify system design. Note that data will appear at the SOUT pin not only on the rising edge of SCLK but also after an access time of  $t_{a(\text{RSO})}$  from  $\overline{\text{RAS}}$  high during a parallel load of the shift registers.

**SIN and SOUT**

Data is shifted in through the SIN pin and is shifted out through the SOUT pin. The TMS4161 is designed such that it requires 0 ns hold time on SIN as SCLK rises. SOUT is guaranteed not to change for at least 8 ns after SCLK rises. These features make it possible to easily connect TMS4161s together, to allow SOUT to be connected to SIN, and to give external circuitry a full SCLK cycle time to allow manipulation of the serial data. To guarantee proper serial clock sequence after power up, a transfer cycle must be initiated before serial data is applied at SIN.

**$\overline{\text{SOE}}$**

The serial output enable pin controls the impedance of the serial output allowing multiplexing of more than one bank of TMS4161 memories into the same external video circuitry. When  $\overline{\text{SOE}}$  is at a low logic level, SOUT will be enabled and the proper data read out. When  $\overline{\text{SOE}}$  is at a high logic level, SOUT will be disabled and be in the high-impedance state.

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Voltage on any pin except $V_{\text{DD}}$ and data out (see Note 1)	.....	-1.5 V to 10 V
Voltage on $V_{\text{DD}}$ supply and data out with respect to $V_{\text{SS}}$	.....	-1 V to 6 V
Short circuit output current	.....	50 mA
Power dissipation	.....	1 W
Operating free-air temperature range	.....	0°C to 70°C
Storage temperature range	.....	-65°C to 150°C

† Stress beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the "Recommended Operating Conditions" section of this specification is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values in this data sheet are with respect to  $V_{\text{SS}}$ .

Texas Instruments  
MEMORY



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- 65,536 X 1 Organization
- Single +5-V Supply (10% Tolerance)
- JEDEC Standardized Pin-Out in Dual-In-Line Packages
- Upward Pin Compatible with TMS4116 (16K Dynamic RAM)
- First Military Version of 64K DRAM
- Available Temperature Ranges:
  - M . . . -55°C to 125°C
  - S . . . -55°C to 100°C
  - E . . . -40°C to 85°C
  - L . . . 0°C to 70°C
- Long Refresh Period . . . 4 milliseconds
- Low Refresh Overhead Time . . . As Low As 1.8% of Total Refresh Period
- All Inputs, Outputs, Clocks Fully TTL Compatible
- 3-State Unlatched Output
- Common I/O Capability with "Early Write" Feature
- Page-Mode Operation for Faster Access
- Low Power Dissipation
  - Operating . . . 125 mW (TYP)
  - Standby . . . 17.5 mW (TYP)
- Performance Ranges:

	ACCESS TIME ROW ADDRESS (MAX)	ACCESS TIME COLUMN ADDRESS (MAX)	READ OR WRITE CYCLE (MIN)	READ- MODIFY- WRITE CYCLE (MIN)
'4164-12	120 ns	70 ns	230 ns	260 ns
'4164-15	150 ns	85 ns	260 ns	285 ns
'4164-20	200 ns	135 ns	330 ns	345 ns

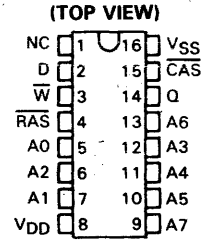
- New SMOS (Scaled-MOS) N-Channel Technology

**description**

The '4164 is a high-speed, 65,536-bit, dynamic random-access memory, organized as 65,536 words of one bit each. It employs state-of-the-art SMOS (scaled MOS) N-channel double-level polysilicon gate technology for very high performance combined with low cost and improved reliability.

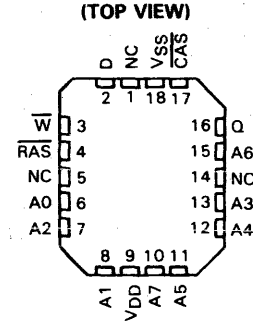
TMS4164 . . . JDL OR NL PACKAGE

SMJ4164 . . . JD PACKAGE

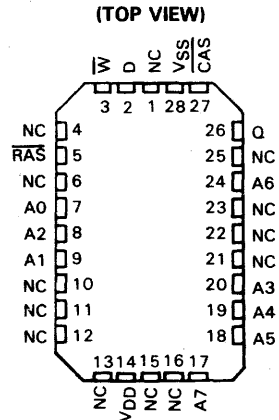


TMS4164 . . . FPL PACKAGE

SMJ4164 . . . FG PACKAGE



SMJ4164 . . . FE PACKAGE



**PIN NOMENCLATURE**

A0-A7	Address Inputs
CAS	Column Address Strobe
D	Data-In
NC	No-Connection
Q	Data-Out
RAS	Row Address Strobe
VDD	+5-V Supply
VSS	Ground
W	Write Enable

**ADVANCE INFORMATION**

**MILITARY PRODUCTS (SMJ) ONLY**

This document contains information on a new product. Specifications are subject to change without notice.

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TEXAS  
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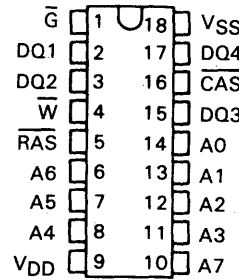


- 16,384 X 4 Organization
- Single +5-V Supply (10% Tolerance)
- Performance Ranges:

	ACCESS TIME ROW ADDRESS (MAX)	ACCESS TIME COLUMN ADDRESS (MAX)	READ OR WRITE CYCLE (MIN)	READ- MODIFY- WRITE CYCLE (MIN)
'4416-15	150 ns	80 ns	260 ns	360 ns
'4416-20	200 ns	120 ns	330 ns	440 ns

- Available Temperature Ranges\*:
  - S . . . -55°C to 100°C
  - E . . . -40°C to 85°C
  - L . . . 0°C to 70°C
- Long Refresh Period . . . 4 milliseconds
- Low Refresh Overhead Time . . . As Low As 1.7% of Total Refresh Period
- All Inputs, Outputs, Clocks Fully TTL Compatible
- 3-State Unlatched Outputs
- Early Write or  $\bar{G}$  to Control Output Buffer Impedance
- Page-Mode Operation for Faster Access
- Low Power Dissipation
  - Operating . . . 130 mW (TYP)
  - Standby . . . 17.5 mW (TYP)
- New SMOS (Scaled-MOS) N-Channel Technology

TMS4416 . . . NL PACKAGE  
SMJ4416 . . . JD PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A0-A7	Address Inputs
$\bar{CAS}$	Column Address Strobe
DQ1-DQ4	Data In/Data Out
$\bar{G}$	Output Enable
$\bar{RAS}$	Row Address Strobe
$V_{DD}$	+5-V Supply
$V_{SS}$	Ground
$\bar{W}$	Write Enable

description

The '4416 is a high-speed, 65,536-bit, dynamic, random-access memory, organized as 16,384 words of 4 bits each. It employs state-of-the-art SMOS (scaled MOS) N-channel double-level polysilicon gate technology for very high performance combined with low cost and improved reliability.

The '4416 features  $\bar{RAS}$  access times to 150 ns maximum. Power dissipation is 130 mW typical operating, 17.5 mW typical standby.

New SMOS technology permits operation from a single +5-V supply, reducing system power supply and decoupling requirements, and easing board layout.  $I_{DD}$  peaks have been reduced to 60 mA typical, and a -1-V input voltage undershoot can be tolerated, minimizing system noise considerations. Input clamp diodes are used to ease system design.

Refresh period is extended to 4 milliseconds, and during this period each of the 256 rows must be strobed with  $\bar{RAS}$  in order to retain data.  $\bar{CAS}$  can remain high during the refresh sequence to conserve power.

All inputs and outputs, including clocks, are compatible with Series 54/74 TTL. All address lines and data-in are latched on chip to simplify system design. Data-out is unlatched to allow greater system flexibility.

\* M temperature range (-55°C to 125°C) to be available in future.

PRODUCT PREVIEW

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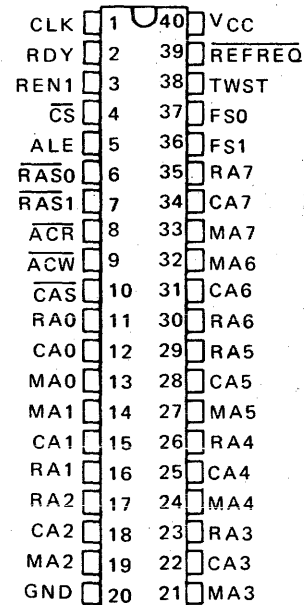
This document contains information on a product under development. Texas Instruments reserves the right to change or discontinue this product without notice.



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- Controls Operation of 8K/16K/32K/64K Dynamic RAMs
- Creates Static RAM Appearance
- One Package Contains Address Multiplexer, Refresh Control, and Timing Control
- Directly Addresses and Drives Up to 256K Bytes of Memory Without External Drivers
- Operates from Microprocessor Clock
  - No Crystals, Delay Lines, or RC Networks
  - Eliminates Arbitration Delays
- Refresh May Be Internally or Externally Initiated
- Versatile
  - Strap-Selected Refresh Rate
  - Synchronous, Predictable Refresh
  - Selection of Distributed, Transparent, and Cycle-Steal Refresh Modes
  - Interfaces Easily to Popular Microprocessors
- Strap-Selected Wait State Generation for Microprocessor/Memory Speed Matching
- Ability to Synchronize or Interleave Controller with the Microprocessor System (Including Multiple Controllers)
- Three-State Outputs Allow Multiport Memory Configuration
- Performance Ranges of 150 ns/200 ns/250 ns

TMS4500A . . . NL PACKAGE  
(TOP VIEW)



**description**

The TMS4500A is a monolithic DRAM system controller designed to provide address multiplexing, timing, control and refresh/access arbitration functions to simplify the interface of dynamic RAMs to microprocessor systems.

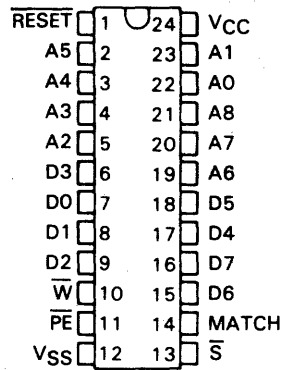
The controller contains a 16-bit multiplexer that generates the address lines for the memory device from the 16 system address bits and provides the strobe signals required by the memory to decode the address. An 8-bit refresh counter generates the 256-row addresses required for refresh.

A refresh timer is provided that generates the necessary timing to refresh the dynamic memories and assure data retention.

The TMS4500A also contains refresh/access arbitration circuitry to resolve conflicts between memory access requests and memory refresh cycles. The TMS4500A is offered in a 40-pin, 600-mil dual-in-line plastic package and is guaranteed for operation from 0°C to 70°C.

- Fast Address to Match Valid Delay - Two Speed Ranges: 45 ns, 55 ns
- 512 X 9 Internal RAM
- 300-Mil 24-Pin Ceramic Side Brazed Package
- Max Power Dissipation: 660 mW
- On-Chip Parity Generation and Checking
- Parity Error Output/Force Parity Error Input
- On-Chip Address/Data Comparator
- Asynchronous, Single-Cycle Reset
- Easily Expandable
- Fully Static, TTL Compatible
- Reliable SMOS (Scaled NMOS) Technology

TMS2150 . . . JDL PACKAGE  
(TOP VIEW)



**description**

The 8-bit-slice cache address comparator consists of a high-speed 512 X 9 static RAM array, parity generator, and parity checker, and 9-bit high-speed comparator. It is fabricated using N-channel silicon gate technology for high speed and simple interface with MOS and bipolar TTL circuits. The cache address comparator is easily cascadable for wider tag addresses or deeper tag memories. Significant reductions in cache memory component count, board area, and power dissipation can be achieved with this device.

When  $\bar{S}$  is low and  $\bar{W}$  is high, the cache address comparator compares the contents of the memory location addressed by A0-A8 with the data on D0-D7 plus generated parity. An equality is indicated by a high level on the MATCH output. A low-level output from  $\bar{PE}$  signifies a parity error in the internal RAM data.  $\bar{PE}$  is an N-channel open-drain output for easy OR-tieing. During a write cycle ( $\bar{S}$  and  $\bar{W}$  low), data on D0-D7 plus generated even parity are written in the 9-bit memory location addressed by A0-A8. Also during write, a parity error may be forced by holding  $\bar{PE}$  low.

A  $\bar{RESET}$  input is provided for initialization. When  $\bar{RESET}$  goes low, all 512 X 9 RAM locations will be cleared and the MATCH output will be forced high.

The cache address comparator operates from a single +5 V supply and is offered in a 24-pin 300-mil side brazed package. The device is fully TTL compatible and is guaranteed to operate from 0°C to 70°C.

**MATCH OUTPUT DESCRIPTION**

MATCH =  $V_{OH}$  if: [A0-A8] = D0-D7 + parity,  
 or:  $\bar{RESET} = V_{IL}$ ,  
 or:  $\bar{S} = V_{IH}$ ,  
 or:  $\bar{W} = V_{IL}$

MATCH =  $V_{OL}$  if: [A0-A8]  $\neq$  D0-D7 + parity,  
 with  $\bar{RESET} = V_{IH}$ ,  
 $\bar{S} = V_{IL}$ , and  $\bar{W} = V_{IH}$

**FUNCTION TABLE**

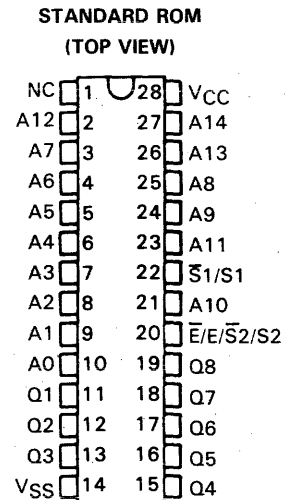
OUTPUT		FUNCTION DESCRIPTION
MATCH	$\bar{PE}$	
L	L	Parity Error
L	H	Not Equal
H	L	Undefined Error
H	H	Equal

Where  $\bar{S} = V_{IL}$ ,  $\bar{W} = V_{IH}$ ,  $\bar{RESET} = V_{IH}$

- 32,768 X 8 Organization
- Fully Static (No clocks, No Refresh)
- All Inputs and Outputs TTL Compatible
- Single 5-V Power Supply
- Optional Power Down or Chip Select
- 64K Bank Select Option
- Maximum Access Time from Address or Power Down:
 

TMS47256-25	250 ns
TMS47256-35	350 ns
TMS47256-45	450 ns
- Worst Case Active Power Dissipation . . . 330 mW
- Worst Case Standby Power Dissipation . . . 66 mW

TMS47256 . . . JL OR NL PACKAGE†



† The package for the bank select ROM is shown on page 2.

**description**

The TMS47256 is a 262,144-bit read-only memory organized as 32,768 words of 8-bit length. This makes the TMS47256 ideal for microprocessor based systems. The device is fabricated using N-channel silicon-gate technology for high speed and simple interface with bipolar circuits.

There are two versions of the TMS47256: the standard ROM with options on chip selects and power down, and the bank select ROM with similar options. The operation section of this data sheet describes both versions.

The TMS47256 is fully compatible with Series 74, 74S, or 74LS TTL. The data outputs are three-state for OR-tieing multiple devices on a common bus. Pins 20 and 22 are mask-programmable, providing additional system flexibility. The data is always available, it is not dependent on external clocking of pins 20 and 22.

The TMS47256 is designed for high-density fixed-memory applications such as logic function generation and microprogramming. It is pin compatible with TI's full line of ROMs and EPROMs.

This ROM is supplied in 28-pin dual-in-line plastic (NL suffix) or ceramic (JL suffix) packages designed for insertion in mounting-hole rows on 600 mil centers. The device is designed for operation from 0°C to 70°C.

**operation, standard ROM**

**address (A0-A14)**

The address-valid interval determines the device cycle time. The 15-bit positive-logic address is decoded on-chip to select one of 32,768 words of 8-bit length in the memory array. A0 is the least-significant bit and A14 the most-significant bit of the word address.

**chip select ( $\bar{S}1$  or S1)**

Pin 22 can be programmed during mask fabrication to be active with either a high- or low-level input. When the signal

PIN NOMENCLATURE	
STANDARD ROM	
A0-A14	Addresses
$\bar{E}/E/\bar{S}2/S2$	Chip Enable/Power Down or Chip Select
NC	No Connection
Q1-Q8	Data Out
$\bar{S}1/S1$	Chip Select
VCC	+5-V Supply
VSS	Ground

**PRODUCT PREVIEW**

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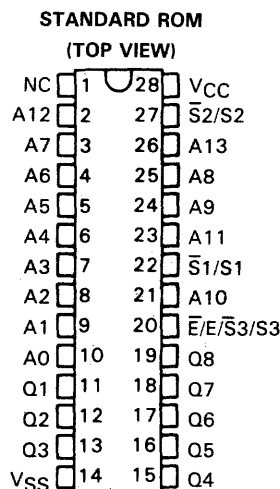
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- 16,384 X 8 Organization
- Fully Static (No Clocks, No Refresh)
- All Inputs and Outputs TTL Compatible
- Single 5-V Power Supply
- Optional Power Down or Chip Select
- 64K Bank Select Option
- Maximum Access Time from Address or Power Down:
 

TMS47128-25	250 ns
TMS47128-35	350 ns
TMS47128-45	450 ns
- Worst Case Active Power Dissipation  
... 330 mW
- Worst Case Standby Power Dissipation  
... 66 mW

TMS47128 . . . JL OR NL PACKAGE†



† The package for the bank select ROM is shown on page 2.

**description**

The TMS47128 is a 131,072-bit read-only memory organized as 16,384 words of 8-bit length. This makes the TMS47128 ideal for microprocessor based systems. The device is fabricated using N-channel silicon-gate technology for high speed and simple interface with bipolar circuits.

There are two versions of the TMS47128: the standard ROM with options on chip selects and power down, and the bank select ROM with similar options. The operation section of this data sheet describes both versions

The TMS47128 is fully compatible with Series 74, 74S, or 74LS TTL. The data outputs are three-state for OR-tying multiple devices on a common bus. Pins 20, 22, and 27 are mask-programmable, providing additional system flexibility. The data is always available, it is not dependent on external clocking of pins 20, 22, or 27.

The TMS47128 is designed for high-density fixed-memory applications such as logic function generation and microprogramming. It is pin compatible with TI's full line of ROMs and EPROMs.

This ROM is supplied in 28-pin dual-in-line plastic (NL suffix) or ceramic (JL suffix) packages designed for insertion in mounting-hole rows on 600 mil centers. The device is designed for operation from 0°C to 70°C.

PIN NOMENCLATURE	
STANDARD ROM	
A0-A13	Addresses
$\bar{E}/E/\bar{S}3/S3$	Chip Enable/Power Down or Chip Select
NC	No Connection
Q1-Q8	Data Out
$\bar{S}1/S1, \bar{S}2/S2$	Chip Selects
VCC	+ 5-V Supply
VSS	Ground

**operation, standard ROM**

**address (A0-A13)**

The address-valid interval determines the device cycle time. The 14-bit positive-logic address is decoded on-chip to select one of 16,384 words of 8-bit length in the memory array. A0 is the least-significant bit and A13 the most-significant bit of the word address.

**chip select ( $\bar{S}1$  or S1 and  $\bar{S}2$  or S2)**

Pins 22 and 27 can be programmed during mask fabrication to be active with either a high- or a low-level input. When

**PRODUCT PREVIEW**

This document contains information on a product under development. Texas Instruments reserves the right to change or discontinue this product without notice.

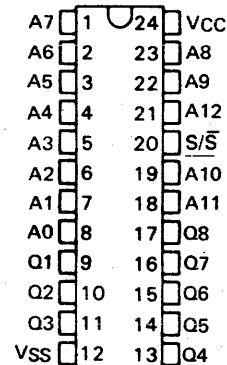
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- 8192 X 8 Organization
- Fully Static (No Clocks, No Refresh)
- All Inputs and Outputs TTL Compatible
- Single 5-V Power Supply
- Maximum Access Time From Address:
  - TMS4764-30 300 ns
  - TMS4764-35 350 ns
  - TMS4764-45 450 ns
- Typical Active Power Dissipation
  - ... 275 mW

TMS4764 . . . JL OR NL PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A0 - A12	Addresses
Q1 - Q8	Data Out
S/ $\bar{S}$	Chip Select
VCC	+5-V Supply
VSS	Ground

**description**

The TMS4764 is a 65,536-bit read-only memory organized as 8192 words of 8-bit length. This makes the TMS4764 ideal for microprocessor based systems. The device is fabricated using N-channel silicon-gate technology for high speed and simple interface with bipolar circuits.

All inputs can be driven directly by Series 74 TTL circuits without the use of any internal pull-up resistor. Each output can drive two Series 74 or 74S loads without external resistors. The data outputs are three-state for OR-tieing multiple devices on a common bus. Pin 20 is programmable, providing additional system flexibility. The data is always available, it is not dependent on external clocking of pin 20.

The TMS4764 is designed for high-density fixed-memory applications such as logic function generation and microprogramming. It is pin compatible with TI's full line of ROMs and EPROMs.

This ROM is supplied in 24-pin dual-in-line-plastic (NL suffix) or ceramic (JL suffix) packages designed for insertion in mounting-hole rows on 600-mil centers or chip on board. The device is designed for operation from 0°C to 70°C.

**operation**

**address (A0 - A12)**

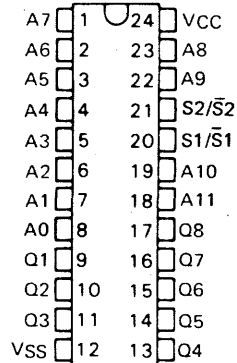
The address-valid interval determines the device cycle time. The 13-bit positive-logic address is decoded on-chip to select one of 8192 words of 8-bit length in the memory array. A0 is the least-significant bit and A12 the most-significant bit of the word address.

**chip select (S or  $\bar{S}$ )**

Pin 20 can be programmed during mask fabrication to be active with either a high- or a low-level input. When the signal is active, all eight outputs are enabled and the eight-bit addressed word can be read. When the signal is not active, all eight outputs are in a high-impedance state.

- 4096 X 8 Organization
- All Inputs and Outputs TTL Compatible
- Fully Static (No Clocks, No Refresh)
- Single 5-V Power Supply
- Maximum Access Time From Address:  
TMS4732-30 300 ns  
TMS4732-35 350 ns  
TMS4732-45 450 ns
- Typical Power Dissipation . . . 275 mW
- 3-State Outputs for OR-Ties
- Pin-Compatible with TMS2532 EPROM
- Two Output Enable Controls for Chip Select Flexibility

TMS4732 . . . JL OR NL PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A0 - A11	Addresses
Q1 - Q8	Data Out
S1/S1-bar, S2/S2-bar	Chip Selects
VCC	+5-V Supply
VSS	Ground

**description**

The TMS4732 is a 32,768-bit read-only memory organized as 4096 words of 8-bit length. This makes the TMS4732 ideal for microprocessor based systems. The device is fabricated using N-channel silicon-gate technology for high speed and simple interface with bipolar circuits.

All inputs can be driven directly by Series 74 TTL circuits without the use of any internal pull-up resistor. Each output can drive one Series 74 or 74S load without external resistors. The data outputs are three-state for OR-tieing multiple devices on a common bus. Two chip-select controls allow data to be read. These controls are programmable, providing additional system decode flexibility. The data is always available, it is not dependent on external clocking of the control pins.

The TMS4732 is designed for high-density fixed-memory applications such as logic function generation and microprogramming. The part is pin compatible with the TMS2532 4096 x 8 EPROM, which aids in prototyping and code verification.

This ROM is supplied in 24-pin dual-in-line-plastic (NL suffix) or ceramic (JL suffix) packages designed for insertion in mounting-hole rows on 600-mil centers or chip on board. The device is designed for operation from 0°C to 70°C.

**operation**

**address (A0 - A11)**

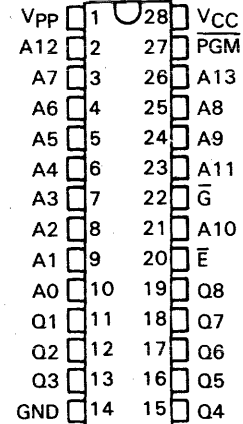
The address-valid interval determines the device cycle time. The 12-bit positive-logic address is decoded on-chip to select one of 4096 words of 8-bit length in the memory array. A0 is the least-significant bit and A11 the most-significant bit of the word address.

**chip select/output enable (pins 20 and 21)**

Each of these pins can be programmed during mask fabrication to be active with either a high- or a low-level input. When both signals are active, all eight outputs are enabled and the eight-bit addressed word can be read. When either signal is not active, all eight outputs are in a high-impedance state.

- 16,384 X 8 Organization
- Single +5-V Power Supply
- Pin Compatible with TMS2764 EPROM
- All Inputs Are TTL Compatible
- Max Access/Min Cycle Time:
  - TMS27128-20      200 ns
  - TMS27128-25      250 ns
  - TMS27128-30      300 ns
  - TMS27128-45      450 ns
- Low Power Dissipation
  - 500 mW (Typical)
- JEDEC Approved Pinout

TMS27128 . . . JL PACKAGE  
(TOP VIEW)



**description**

The TMS27128 is an ultraviolet light-erasable, electrically programmable read-only memory. It has 131,072 bits organized as 16,384 words of 8-bit length. The TMS27128 only requires a single 5-volt power supply with a tolerance of  $\pm 5\%$ . The TMS27128-20 provides an access time of 200 ns, which is compatible with high-speed microprocessors.

PIN NOMENCLATURE	
A0-A13	Addresses
$\bar{E}$	Chip Enable/Power Down
$\bar{G}$	Output Enable
GND	Ground
PGM	Program
Q1-Q8	Outputs
VCC	+5-V Power Supply
Vpp	+21-V Power Supply

The TMS27128 provides dual output control lines: Output Enable ( $\bar{G}$ ) and Chip Enable/Power Down ( $\bar{E}$ ). This feature allows the  $\bar{G}$  control line to eliminate bus contention in multibus microprocessor systems. The TMS27128 has a standby mode that reduces the maximum power dissipation from 750 mW to 225 mW when the device is placed on standby.

This EPROM is supplied in a 28-pin dual-in-line ceramic package (JL suffix). It is pin compatible with the TMS2764 EPROM and is designed for operation from 0°C to 70°C.

**operation**

The five modes of operation for the TMS27128 are listed in the following table.

FUNCTION (PINS)	MODE				
	Read	Power Down (Standby)	Program	Program Verification	Inhibit Programming
PGM (27)	V <sub>IH</sub>	X	V <sub>IL</sub>	V <sub>IH</sub>	X
VPP (1)	5 V	5 V	21 V	21 V	21 V
VCC (28)	VCC	VCC	5 V	5 V	5 V
$\bar{E}$ (20)	V <sub>IL</sub>	V <sub>IH</sub>	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>
$\bar{G}$ 22	V <sub>IL</sub>	X	V <sub>IH</sub>	V <sub>IL</sub>	X
Q1-Q8 (11 to 13, 15 to 19)	Q	HI-Z	D	Q	HI-Z

X = V<sub>IL</sub> or V<sub>IH</sub>

**PRODUCT PREVIEW**

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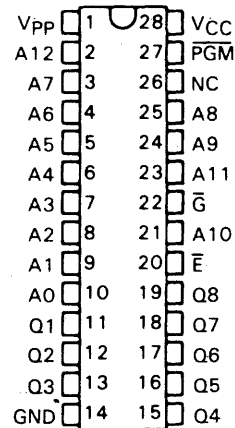


- Organization . . . 8192 X 8
- Single +5-V Power Supply
- Pin Compatible with TMS2732A EPROM
- All Inputs and Outputs are TTL Compatible
- Performance Ranges:

	MAX ACCESS/ MIN CYCLE TIME	VOLTAGE TOLERANCE
TMS2764H-20	200 ns	± 5%
TMS2764H-25	250 ns	± 5%
TMS2764H-30	300 ns	± 10%
TMS2764H-45	450 ns	± 10%

- Low Active Current . . . 100 mA (Max)
- JEDEC Approved Pinout
- 21-V Power Supply Required for Programming
- Fast Programming Algorithm Compatible
- N-Channel Silicon-Gate Technology
- 8-Bit Output for Use in Microprocessor-Based Systems
- Static Operation (No Clocks, No Refresh)

TMS2764H . . . JL PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A0-A12	Addresses
$\bar{E}$	Chip Enable
NC	No Connection
$\bar{G}$	Output Enable
Q1-Q8	Outputs
$\overline{PGM}$	Program

description

The TMS2764H is an ultraviolet light-erasable, electrically programmable read-only memory. It has 65,536 bits organized as 8,192 words of 8-bit length. The TMS2764H-20 only requires a single 5-volt power supply with a tolerance of ±5%, and has a maximum access time of 200 ns. This access time is compatible with high-performance microprocessors.

The TMS2764H provides two output control lines: Output Enable ( $\bar{G}$ ) and Chip Enable ( $\bar{E}$ ). This feature allows the  $\bar{G}$  control line to eliminate bus contention in multibus microprocessor systems. The TMS2764H has a power-down mode that reduces maximum active current from 100 mA to 35 mA when the device is placed on standby.

This EPROM is supplied in a 28-pin, 600-mil dual-in-line ceramic package and is designed for operation from 0°C to 70°C.

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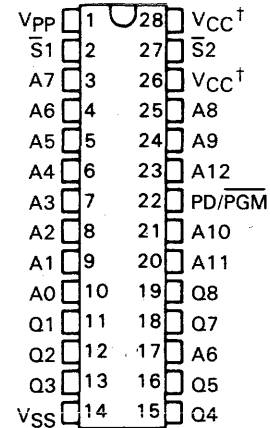
- Organization . . . 8K X 8
- Single +5-V Power Supply
- Pin Compatible with Existing ROMs and EPROMs (8K, 16K, 32K, and 64K)
- All Input/Outputs Fully TTL Compatible
- Static Operation (No Clocks, No Refresh)
- Max Access/Min Cycle Time:
  - TMS2564-35 . . . 350 ns
  - TMS2564-45 . . . 450 ns
  - SMJ2564-45 . . . 450 ns
- 8-Bit Output for Use in Microprocessor-Based Systems
- N-Channel Silicon-Gate Technology
- 3-State Output Buffers
- Guaranteed DC Noise Immunity with Standard TTL Loads
- No Pull-Up Resistors Required
- Low Power Dissipation:
  - Active . . . 400 mW Typical
  - Standby . . . 75 mW Typical
- Available in Full Military Temperature Range Version (SMJ2564)

**description**

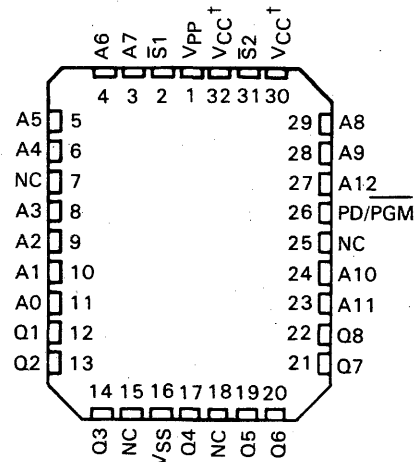
The '2564 is a 65,536-bit ultraviolet-light-erasable, electrically-programmable read-only memory. This device is fabricated using N-channel silicon-gate technology for high-speed and simple interface with MOS and bipolar circuits. All inputs (including program data inputs) can be driven by Series 54/74 TTL circuits without the use of external resistors. The data outputs are three-state for connecting multiple devices to a common bus.

The TMS2564 is offered in a dual-in-line ceramic package (JL or JDL suffix) rated for operation from 0°C to 70°C. The SMJ2564 is offered in a 28-pin dual-in-line ceramic package (J) and a leadless ceramic chip carrier (FE), rated for operation from -55°C to 125°C. The J package is designed for insertion in mounting-hole rows on 600-mil (15.2 mm) centers, whereas the FE package is intended for surface mounting on solder pads on 0.050-inch (1.27 mm) centers. The FE package offers a three-layer rectangular chip carrier with dimensions 0.450 x 0.550 x 0.100 (11.43 x 13.97 x 2.54 mm).

TMS2564 . . . JL OR JDL PACKAGE  
SMJ2564 . . . J PACKAGE  
(TOP VIEW)



SMJ2564 . . . FE PACKAGE  
(TOP VIEW)



† Connected internally, VCC need be supplied to only one of these two pins.

PIN NOMENCLATURE	
A(N)	Address Inputs
NC	No Connection
PD/PGM	Power Down/Program
Q(N)	Input/Output
S(N)	Chip Selects
VCC	+5-V Power Supply
VPP	+25-V Power Supply
VSS	0-V Ground

ADVANCE INFORMATION  
MILITARY PRODUCTS (SMJ) ONLY

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# TMS2732A

## 32,768-BIT ERASABLE PROGRAMMABLE READ-ONLY MEMORY

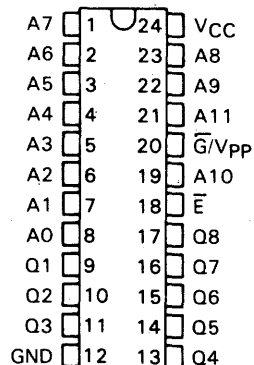
AUGUST 1983

- Organization . . . 4096 X 8
- Single +5-V Power Supply
- All Inputs and Outputs Are TTL Compatible
- Performance Ranges:

	ACCESS TIME (MAX)	CYCLE TIME (MIN)
TMS2732A-30	300 ns	300 ns
TMS2732A-35	350 ns	350 ns
TMS2732A-45	450 ns	450 ns

- Low Standby Power Dissipation . . .  
158 mW (Maximum)
- JEDEC Approved Pinout . . . Industry Standard
- 21-V Power Supply Required for Programming
- N-Channel Silicon-Gate Technology
- 8-Bit Output for Use in Microprocessor Based Systems
- Static Operation (No Clocks, No Refresh)

TMS2732A . . . JL PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A0 - A11	Addresses
$\bar{E}$	Chip Enable
$\bar{G}/V_{pp}$	Output Enable/ +21 V
Q1 - Q8	Outputs
VCC	+5-V Power Supply

### description

The TMS2732A is an ultraviolet light-erasable, electrically programmable read-only memory. It has 32,768 bits organized as 4,096 words of 8-bit length. The TMS2732A only requires a single 5-volt power supply with a tolerance of  $\pm 5\%$ .

The TMS2732A provides two output control lines: Output Enable ( $\bar{G}$ ) and Chip Enable ( $\bar{E}$ ). This feature allows the  $\bar{G}$  control line to eliminate bus contention in multibus microprocessor systems. The TMS2732A has a power-down mode that reduces maximum power dissipation from 657 mW to 158 mW when the device is placed on standby.

This EPROM is supplied in a 24-pin dual-in-line ceramic package and is designed for operation from 0°C to 70°C.

### operation

The six modes of operation for the TMS2732A are listed in the following table.

FUNCTION (PINS)	MODE					
	Read	Deselect	Power Down (Standby)	Program	Program Verification	Inhibit Programming
$\bar{E}$ (18)	$V_{IL}$	Don't Care	$V_{IH}$	$V_{IL}$	$V_{IL}$	$V_{IH}$
$\bar{G}/V_{pp}$ (20)	$V_{IL}$	$V_{IH}$	Don't Care	21 V	$V_{IL}$	21 V
VCC (24)	5 V	5 V	5 V	5 V	5 V	5 V
Q1-Q8 (9 to 11, 13 to 17)	Q	HI-Z	HI-Z	D	Q	HI-Z

### PRODUCT PREVIEW

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- Organization . . . 4096 X 8
- Single +5-V Power Supply
- Pin Compatible with Existing ROMs and EPROMs (8K, 16K, 32K, and 64K)
- JEDEC Standard Pinout
- All Inputs/Outputs Fully TTL Compatible
- Static Operation (No Clocks, No Refresh)
- Max Access/Min Cycle Time:
 

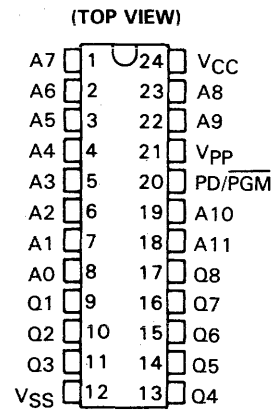
'2532-30	300 ns
'2532-35	350 ns
'2532-45	450 ns
- 8-Bit Output for Use in Microprocessor-Based Systems
- N-Channel Silicon-Gate Technology
- 3-State Output Buffers
- Low Power Dissipation:
  - Active . . . 400 mW Typical
  - Standby . . . 50 mW Standby
- Guaranteed DC Noise Immunity with Standard TTL Loads
- No Pull-Up Resistors Required
- Available in Full Military Temperature Range Version (SMJ2532)

**description**

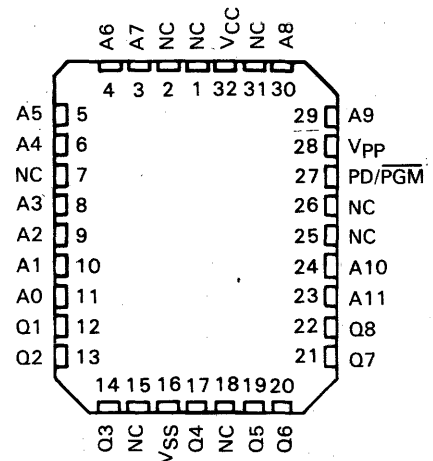
The '2532 series are 32,768-bit, ultraviolet-light-erasable, electrically-programmable read-only memories. These devices are fabricated using N-channel silicon-gate technology for high speed and simple interface with MOS and bipolar circuits. All inputs (including program data inputs) can be driven by Series 54/74 TTL circuits without the use of external pull-up resistors, and each output can drive one Series 54/74 TTL circuit without external resistors. The data outputs are three-state for connecting multiple devices to a common bus. The TMS2532 series are plug-in compatible with the TMS4732 32K ROM.

The TMS2532's are offered in a dual-in-line ceramic package (JL suffix), rated for operation from 0°C to 70°C. The SMJ' devices are offered in a 24-pin dual-in-line ceramic package (J) and in a 32-pad leadless ceramic chip carrier (FE). The J package is designed for insertion in mounting-hole rows on 600-mil (15.2 mm) centers, whereas the FE package is intended for surface mounting on solder pads on 0.050-inch (1.27 mm) centers. The FE package offers a three-layer rectangular chip carrier with dimensions 0.450 x 0.550 x 0.100 (11.43 x 13.97 x 2.54 mm).

TMS2532 . . . JL PACKAGE  
SMJ2532 . . . J PACKAGE



SMJ2532 . . . FE PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A(N)	Address Inputs
NC	No Internal Connection
PD/PGM	Power Down/Program
Q(N)	Data Outputs
VCC	+5-V Power Supply
Vpp	+25-V Supply
VSS	0-V Ground

- Organization . . . 2048 X 8
- Single +5-V Power Supply
- Pin Compatible with Existing ROMs and EPROMs (16K, 32K, and 64K)
- JEDEC Standard Pinout
- All Inputs/Outputs Fully TTL Compatible
- Static Operation (No Clocks, No Refresh)
- Max Access/Min Cycle Time:
 

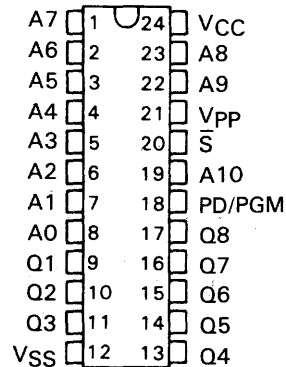
'2516-35	350 ns
'2516-45	450 ns
- 8-Bit Output for Use in Microprocessor-Based Systems
- N-Channel Silicon-Gate Technology
- 3-State Output Buffers
- 40% Lower Power
 

Active	285 mW Typical
Standby	50 mW Typical
- Guaranteed DC Noise Immunity with Standard TTL Loads
- No Pull-Up Resistors Required
- Available in Full Military Temperature Range Version (SMJ2516)

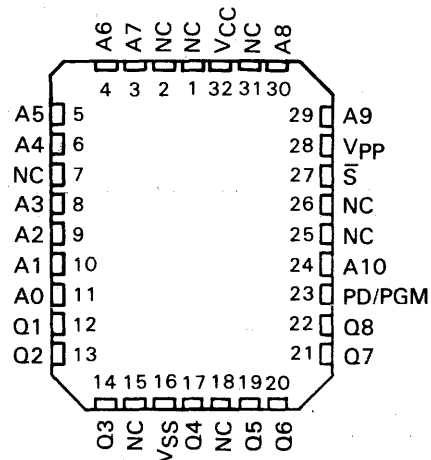
**description**

The '2516 series are 16,384-bit, ultraviolet-light erasable, electrically-programmable read-only memories. These devices are fabricated using N-channel silicon-gate technology for high speed and simple interface with MOS and bipolar circuits. All inputs (including program data inputs) can be driven by Series 54/74 TTL circuits without the use of external pull-up resistors, and each output can drive one Series 54/74 TTL circuit without external resistors. The data outputs are three-state for connecting multiple devices to a common bus. The '2516 is plug-in compatible with the '4016 16K static RAM.

TMS2516 . . . JL PACKAGE  
SMJ2516 . . . J PACKAGE  
(TOP VIEW)



SMJ2516 . . . FE PACKAGE  
(TOP VIEW)



PIN NOMENCLATURE	
A(N)	Address Inputs
NC	No Internal Connection
PD/PGM	Power Down/Program
Q(N)	Data Outputs
$\bar{S}$	Chip Select
VCC	+5-V Power Supply
VPP	+25-V Power Supply
VSS	0-V Ground

- 2048 X 8 Organization
- All Inputs and Outputs Fully TTL Compatible
- Static Operation (No Clocks, No Refresh)
- Performance Ranges:

	ACCESS TIME (MAX)	CYCLE TIME (MIN)
TMS2716-30	300 ns	300 ns
TMS2716-45	450 ns	450 ns

- 3-State Outputs for OR-Ties
- N-Channel Silicon-Gate Technology
- 8-Bit Output for Use in Microprocessor-Based Systems
- Low Power . . . 315 mW (Typical)

**description**

The TMS2716 is an ultra-violet light-erasable, electrically programmable read-only memory. It has 16,384 bits organized as 2048 words of 8-bit length. The device is fabricated using N-channel silicon-gate technology for high-speed and simple interface with MOS and bipolar circuits. All inputs (including program data inputs) can be driven by Series 74 circuits without the use of external pull-up resistors and each output can drive one Series 74 or 74LS TTL circuit without external resistors. The TMS2716 guarantees 250 mV dc noise immunity in the low state. Data outputs are three-state for OR-tying multiple devices on a common bus. The TMS2716 is plug-in compatible with the TMS2708 and the TMS27L08. Pin compatible mask programmed ROMs are available for large volume requirements.

This EPROM is designed for high-density fixed-memory applications where fast turn arounds and/or program changes are required. It is supplied in a 24-pin dual-in-line cerpak (JL suffix) package designed for insertion in mounting-hole rows on 600-mil (15,2 mm) centers. It is designed for operation from 0°C to 70°C.

**operation (read mode)**

**address (A0-A10)**

The address-valid interval determines the device cycle time. The 11-bit positive-logic address is decoded on-chip to select one of 2048 words of 8-bit length in the memory array. A0 is the least-significant bit and A10 most-significant bit of the word address.

**chip select, program [ $\bar{S}$  (PGM)]**

When the chip select is low, all eight outputs are enabled and the eight-bit addressed word can be read. When the chip select is high, all eight outputs are in a high-impedance state.

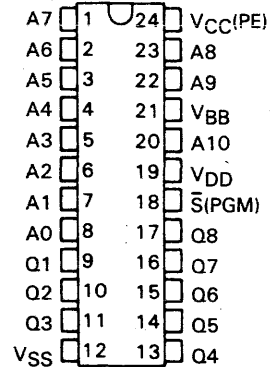
**program**

In the program mode, the chip select feature does not function as pin 18 inputs only the program pulse. The program mode is selected by the V<sub>CC</sub>(PE) pin. Either 0 V or + 12 V on this pin will cause the TMS2716 to assume program cycle.

**data out (Q1-Q8)**

The chip must be selected before the eight-bit output word can be read. Data will remain valid until the address is changed or the chip is deselected. When deselected, the three-state outputs are in a high-impedance state. The outputs will drive TTL circuits without external components.

**TMS2716 . . . JL PACKAGE**  
**(TOP VIEW)**



PIN NOMENCLATURE	
A0-A10	Addresses
Q1-Q8	Data Out
$\bar{S}$ (PGM)	Chip Select (Program)
V <sub>BB</sub>	-5-V Supply
V <sub>CC</sub> (PE)	+5-V Supply (Program Enable)
V <sub>DD</sub>	+12-V Supply
V <sub>SS</sub>	0 V Ground

- 1024 X 8 Organization
- All Inputs and Outputs Fully TTL Compatible
- Static Operation (No Clocks, No Refresh)
- Max Access/Min Cycle Time
 

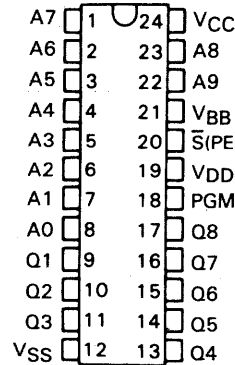
'2708-35	350 ns
'2708-45	450 ns
'27L08-45	450 ns
- 3-State Outputs for OR-Ties
- N-Channel Silicon-Gate Technology
- 8-Bit Output for Use in Microprocessor-Based Systems
- Power Dissipation
 

'27L08	580 mW Max Active
'2708	800 mW Max Active
- 10% Power Supply Tolerance (TMS27L08-45 and all SMJ' versions)
- Plug-Compatible Pin-Outs Allowing Interchangeability/Upgrade to 16K With Minimum Board Change
- Available in Full Military Temperature Range Versions (SMJ2708)

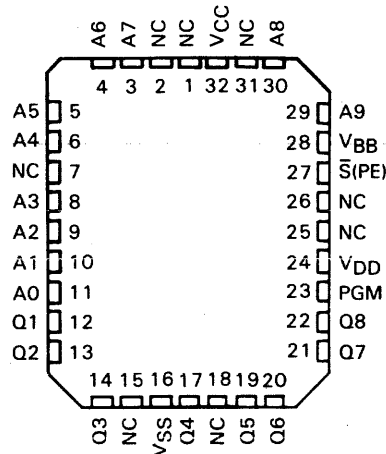
**description**

The '2708-35, '2708-45, and '27L08-45 are ultraviolet light-erasable, electrically programmable read-only memories. They have 8,192 bits organized as 1024 words of 8-bit length. The devices are fabricated using N-channel silicon-gate technology for high speed and simple interface with MOS and bipolar circuits. All inputs (including program data inputs) can be driven by Series 54/74 TTL circuits without the use of external pull-up resistors. Each output can drive one Series 54/74 or 54LS/74LS TTL circuit without external resistors. The '27L08 guarantees 200 mV dc noise immunity in the high state and 250 mV in the low state. The data outputs for the '2708-35, '2708-45, and '27L08-45 are three-state for OR-tying multiple devices on a common bus.

TMS2708 . . . JL PACKAGE  
SMJ2708 . . . J PACKAGE  
(TOP VIEW)



SMJ2708 . . . FE PACKAGE  
(TOP VIEW)



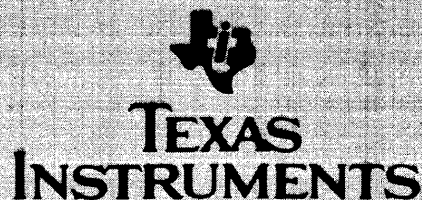
NC - No Connection

**PIN NOMENCLATURE**

A0-A7	Address Inputs
NC	No Connection
PGM	Program
Q1-Q8	Data Out
S(PE)	Chip Select/Program Enable
VBB	-5-V Power Supply
VCC	+5-V Power Supply
VDD	+12-V Power Supply
VSS	0-V Ground

# Bipolar Programmable Read-Only Memories Product Guide

High-Performance  
Schottky PROMs  
Low-Power PROMs





**NUMERICAL INDEX**  
**HIGH-PERFORMANCE SCHOTTKY PROMS**

DEVICE	FUNCTION	AVAILABILITY
TBP18S030	32 X 8 PROM's	NOW
TBP18SA030		NOW
TBP24S10	256 X 4 PROM's	NOW
TBP24SA10		NOW
TBP24S41	1024 X 4 PROM's	NOW
TBP24SA41		NOW
TBP24S81	2048 X 4 PROM's	NOW
TBP24S81-55		NOW
TBP24SA81		NOW
TBP24SA81-55		NOW
TBP28L22	256 X 8 Low-Power PROM's	NOW
TBP28LA22		NOW
TBP28L42	512 X 8 Low-Power PROM's	NOW
TBP28L46		NOW
TBP28L86A	1024 X 8 Low-Power PROM's	▲
TBP28L166	2048 X 8 Low-Power PROM's	NOW
TBP28S42	512 X 8 PROM's	NOW
TBP28SA42		NOW
TBP28S46		NOW
TBP28SA46		NOW
TBP28S86A	1024 X 8 PROM's	NOW
TBP28S86A-50		NOW
TBP28SA86A		NOW
TBP28SA86A-50		NOW
TBP28S165	2048 X 8 PROM's	▲
TBP28S166		NOW
TBP28S2708A	1024 X 8 PROM's	NOW
<p>▲ Planned for future announcement. Contact TI field sales office or authorized TI distributor for status.</p>		

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MEMORY

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**TBP14S10 ('S287)** This product is no longer in production, it is replaced by TBP24S10.

**TBP14SA10 ('S387)** This product is no longer in production, it is replaced by TBP24SA10.

**TBP18S22 ('S471)** This product is no longer in production, it is replaced by TBP28L22.

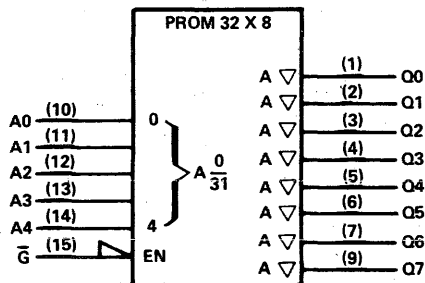
**TBP18SA22 ('S470)** This product is no longer in production, it is replaced by TBP28LA22.

**TBP18S030 ('S288)**

PROGRAMMABLE READ-ONLY MEMORIES

- 32 X 8
- Three-state outputs
- Typical address access time ... 25 ns
- Typical power ... 400 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	Q0	9	Q7
2	Q1	10	A0
3	Q2	11	A1
4	Q3	12	A2
5	Q4	13	A3
6	Q5	14	A4
7	Q6	15	$\bar{G}$
8	GND	16	V <sub>CC</sub>

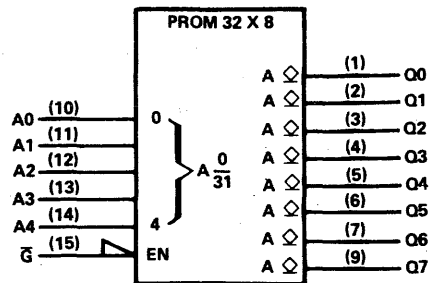
For chip carrier options and information, contact the factory.

**TBP18SA030 ('S188)**

PROGRAMMABLE READ-ONLY MEMORIES

- 32 X 8
- Open-collector outputs
- Typical address access time ... 25 ns
- Typical power ... 400 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	Q0	9	Q7
2	Q1	10	A0
3	Q2	11	A1
4	Q3	12	A2
5	Q4	13	A3
6	Q5	14	A4
7	Q6	15	$\bar{G}$
8	GND	16	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

**TBP18S42 ('S472)** This product is no longer in production, it is replaced by TBP28S42.

**TBP18SA42 ('S473)** This product is no longer in production, it is replaced by TBP28SA42.

**TBP18S46 ('S474)** This product is no longer in production, it is replaced by TBP28S46.

**TBP18SA46 ('S475)** This product is no longer in production, it is replaced by TBP28SA46.

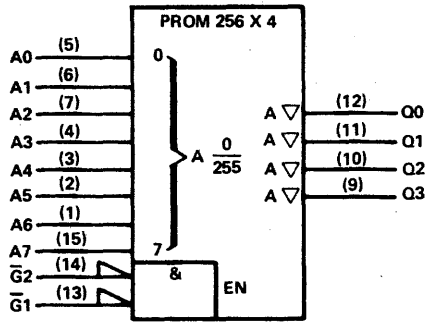
† Pin numbers shown on logic symbols are for J and N packages only.  
nc — no internal connection.

### TBP24S10

**PROGRAMMABLE READ-ONLY MEMORIES**

- 256 X 4
- Three-state outputs
- Typical address access time . . . 35 ns
- Typical select time . . . 20 ns
- Typical power . . . 375 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A6	9	Q3
2	A5	10	Q2
3	A4	11	Q1
4	A3	12	Q0
5	A0	13	G1
6	A1	14	G2
7	A2	15	A7
8	GND	16	VCC

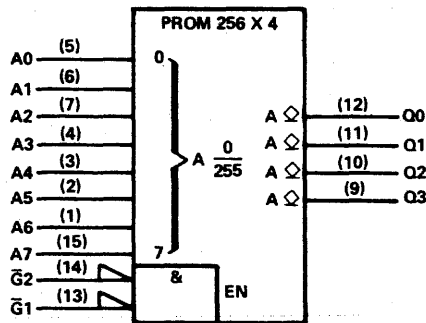
For chip carrier options and information, contact the factory.

### TBP24SA10

**PROGRAMMABLE READ-ONLY MEMORIES**

- 256 X 4
- Open-collector outputs
- Typical address access time . . . 35 ns
- Typical select time . . . 20 ns
- Typical power . . . 375 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A6	9	Q3
2	A5	10	Q2
3	A4	11	Q1
4	A3	12	Q0
5	A0	13	G1
6	A1	14	G2
7	A2	15	A7
8	GND	16	VCC

For chip carrier options and information, contact the factory.

† Pin numbers shown on logic symbols are for J and N packages only.  
nc — no internal connection.

For additional information on other Texas Instruments Military products, refer to the Military section.

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**TEXAS INSTRUMENTS**

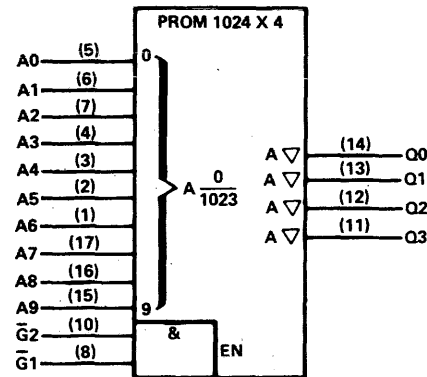
POST OFFICE BOX 225012 • DALLAS, TEXAS 75265

### TBP24S41 ('S476)

#### PROGRAMMABLE READ-ONLY MEMORIES

- 1024 X 4
- Three-state outputs
- Typical address access time . . . 40 ns
- Typical select time . . . 20 ns
- Typical power . . . 475 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A6	10	G2
2	A5	11	Q3
3	A4	12	Q2
4	A3	13	Q1
5	A0	14	Q0
6	A1	15	A9
7	A2	16	A8
8	G1	17	A7
9	GND	18	V <sub>CC</sub>

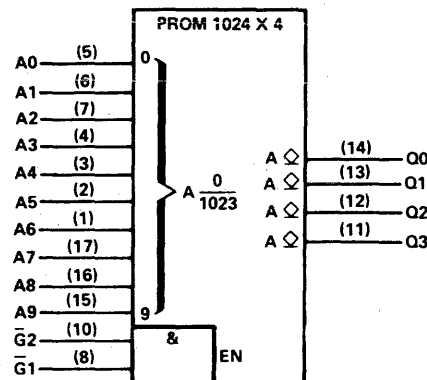
For chip carrier options and information, contact the factory.

### TBP24SA41 ('S477)

#### PROGRAMMABLE READ-ONLY MEMORIES

- 1024 X 4
- Open-collector outputs
- Typical address access time . . . 40 ns
- Typical select time . . . 20 ns
- Typical power . . . 475 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A6	10	G2
2	A5	11	Q3
3	A4	12	Q2
4	A3	13	Q1
5	A0	14	Q0
6	A1	15	A9
7	A2	16	A8
8	G1	17	A7
9	GND	18	V <sub>CC</sub>

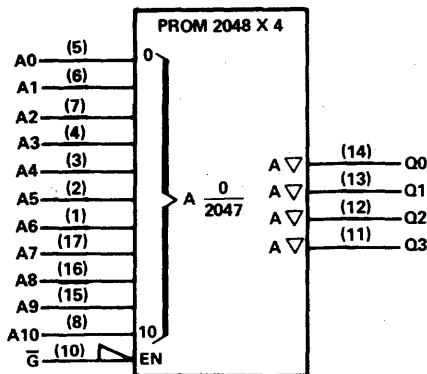
For chip carrier options and information, contact the factory.

### TBP24S81 ('S454) TBP24S81-55

#### PROGRAMMABLE READ-ONLY MEMORIES

- 2048 X 4
- Three-state outputs
- Typical address access time . . . 45 ns
- TBP24S81-55 maximum address access time . . . 55 ns
- Typical select time . . . 20 ns
- Typical power . . . 625 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A6	10	G1
2	A5	11	Q3
3	A4	12	Q2
4	A3	13	Q1
5	A0	14	Q0
6	A1	15	A9
7	A2	16	A8
8	A10	17	A7
9	GND	18	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

† Pin numbers shown on logic symbols are for J and N packages only.  
nc — no internal connection.

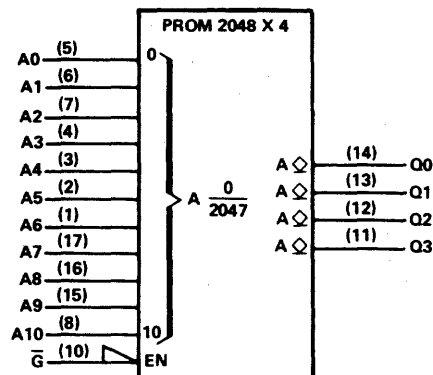
For additional information on other Texas Instruments Military products, refer to the Military section.

**TBP24SA81 ('S455)**  
**TBP24SA81-55**

**PROGRAMMABLE READ-ONLY MEMORIES**

- 2048 X 4
- Open-collector outputs
- Typical address access time . . . 45 ns
- TBP24SA81-55 maximum address access time . . . 55 ns
- Typical select time . . . 20 ns
- Typical power . . . 625 mW

logic symbol†



pin assignments

J, N PACKAGES		
1	A6	10 $\bar{G}$
2	A5	11 Q3
3	A4	12 Q2
4	A3	13 Q1
5	A0	14 Q0
6	A1	15 A9
7	A2	16 A8
8	A10	17 A7
9	GND	18 V <sub>CC</sub>

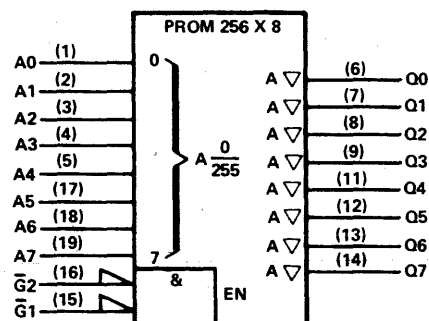
For chip carrier options and information, contact the factory.

**TBP28L22**

**LOW-POWER PROGRAMMABLE READ-ONLY MEMORIES**

- 256 X 8
- Three-state outputs
- Typical address access time . . . 45 ns
- Typical select time . . . 20 ns
- Typical power . . . 375 mW

logic symbol†



pin assignments

J, N PACKAGES		
1	A0	11 Q4
2	A1	12 Q5
3	A2	13 Q6
4	A3	14 Q7
5	A4	15 $\bar{G}$ 1
6	Q0	16 $\bar{G}$ 2
7	Q1	17 A5
8	Q2	18 A6
9	Q3	19 A7
10	GND	20 V <sub>CC</sub>

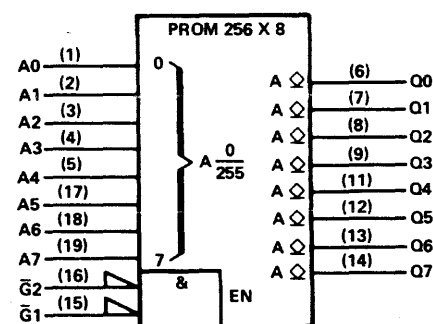
For chip carrier options and information, contact the factory.

**TBP28LA22**

**LOW-POWER PROGRAMMABLE READ-ONLY MEMORIES**

- 256 X 8
- Open-collector outputs
- Typical address access time . . . 45 ns
- Typical select time . . . 20 ns
- Typical power . . . 375 mW

logic symbol†



pin assignments

J, N PACKAGES		
1	A0	11 Q4
2	A1	12 Q5
3	A2	13 Q6
4	A3	14 Q7
5	A4	15 $\bar{G}$ 1
6	Q0	16 $\bar{G}$ 2
7	Q1	17 A5
8	Q2	18 A6
9	Q3	19 A7
10	GND	20 V <sub>CC</sub>

For chip carrier options and information, contact the factory.

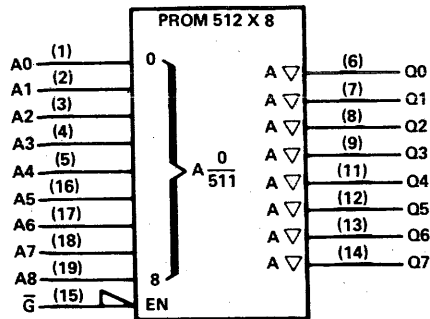
† Pin numbers shown on logic symbols are for J and N packages only.  
nc — no internal connection.

# TBP28L42

## LOW-POWER PROGRAMMABLE READ-ONLY MEMORIES

- 512 X 8
- Three-state outputs
- Typical address access time ... 55 ns
- Typical select time ... 25 ns
- Typical power ... 275 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A0	11	Q4
2	A1	12	Q5
3	A2	13	Q6
4	A3	14	Q7
5	A4	15	$\bar{G}$
6	Q0	16	A5
7	Q1	17	A6
8	Q2	18	A7
9	Q3	19	A8
10	GND	20	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

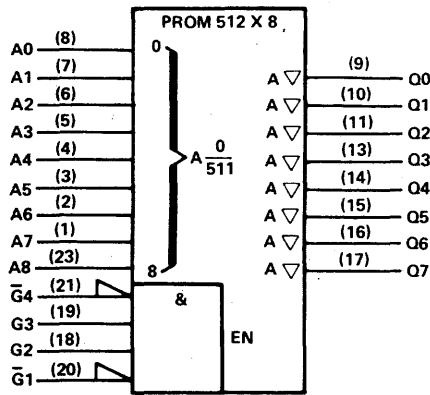
† Pin numbers shown on logic symbols are for J and N packages only.  
nc - no internal connection.

**TBP28L45**  
**TBP28L46**

**LOW-POWER PROGRAMMABLE  
READ-ONLY MEMORIES**

- 512 X 8
- Three-state outputs
- Typical address access time . . . 55 ns
- Typical select time . . . 25 ns
- Typical power . . . 275 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1
9	Q0	21	G4
10	Q1	22	nc
11	Q2	23	A8
12	GND	24	VCC

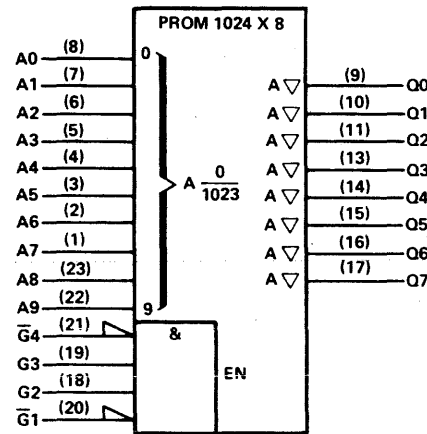
For chip carrier options and information, contact the factory.

**TBP28L85A**  
**TBP28L86A**

**LOW-POWER PROGRAMMABLE  
READ-ONLY MEMORIES**

- 1024 X 8
- Three-state outputs
- Typical address access time . . . 65 ns
- Typical select time . . . 30 ns
- Typical power . . . 275 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1
9	Q0	21	G4
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	VCC

For chip carrier options and information, contact the factory.

**TBP28L86 ('LS478)** This product is no longer in production, it is replaced by TBP28L86A.

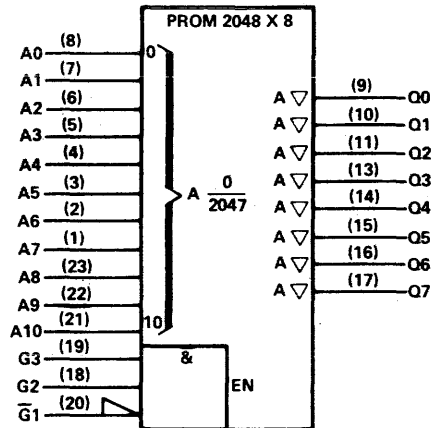
† Pin numbers shown on logic symbols are for J and N packages only.  
nc - no internal connection.

# TBP28L166

## LOW-POWER PROGRAMMABLE READ-ONLY MEMORIES

- 2048 X 8
- Three-state outputs
- Typical address access time  
... 65 ns
- Typical select time ... 30 ns
- Typical power ... 350 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	$\bar{G}1$
9	Q0	21	A10
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	VCC

For chip carrier options and information, contact the factory.

† Pin numbers shown on logic symbols are for J and N packages only.  
nc — no internal connection.

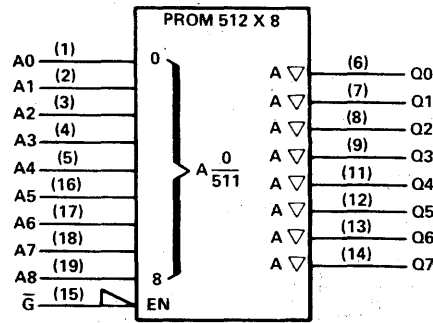


# TBP28S42

## PROGRAMMABLE READ-ONLY MEMORIES

- 512 X 8
- Three-state outputs
- Typical address access time . . . 35 ns
- Typical select time . . . 20 ns
- Typical power . . . 500 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A0	11	Q4
2	A1	12	Q5
3	A2	13	Q6
4	A3	14	Q7
5	A4	15	$\bar{G}$
6	Q0	16	A5
7	Q1	17	A6
8	Q2	18	A7
9	Q3	19	A8
10	GND	20	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

† Pin numbers shown on logic symbols are for J and N packages only.

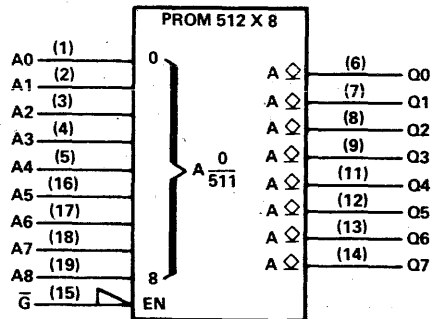
nc – no internal connection.

### TBP28SA42

#### PROGRAMMABLE READ-ONLY MEMORIES

- 512 X 8
- Open-collector outputs
- Typical address access time ... 35 ns
- Typical select time ... 20 ns
- Typical power ... 500 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A0	11	Q4
2	A1	12	Q5
3	A2	13	Q6
4	A3	14	Q7
5	A4	15	G-bar
6	Q0	16	A5
7	Q1	17	A6
8	Q2	18	A7
9	Q3	19	A8
10	GND	20	VCC

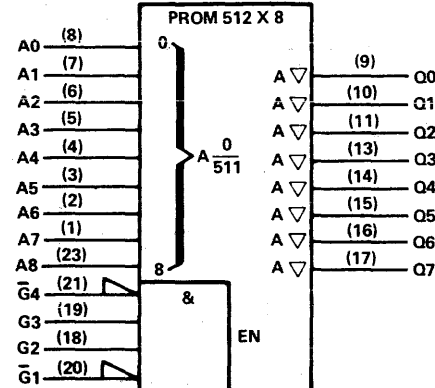
For chip carrier options and information, contact the factory.

### TBP28S45 TBP28S46

#### PROGRAMMABLE READ-ONLY MEMORIES

- 512 X 8
- Three-state outputs
- Typical address access time ... 35 ns
- Typical select time ... 20 ns
- Typical power ... 500 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1-bar
9	Q0	21	G4-bar
10	Q1	22	nc
11	Q2	23	A8
12	GND	24	VCC

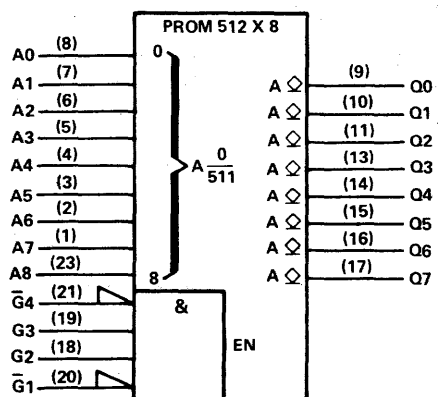
For chip carrier options and information, contact the factory.

### TBP28SA45 TBP28SA46

#### PROGRAMMABLE READ-ONLY MEMORIES

- 512 X 8
- Open-collector outputs
- Typical address access time ... 35 ns
- Typical select time ... 20 ns
- Typical power ... 500 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1-bar
9	Q0	21	G4-bar
10	Q1	22	nc
11	Q2	23	A8
12	GND	24	VCC

For chip carrier options and information, contact the factory.

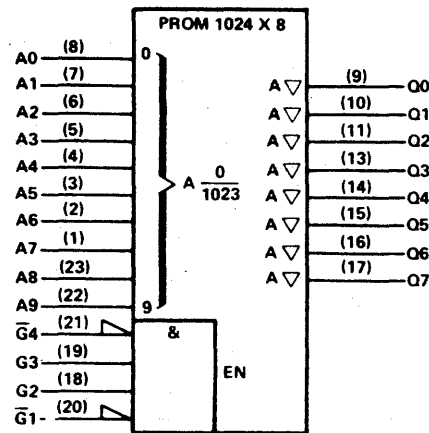
† Pin numbers shown on logic symbols are for J and N packages only.  
nc — no internal connection.

**TBP28S85A**  
**TBP28S85A-50**

**PROGRAMMABLE READ-ONLY MEMORIES**

- 1024 X 8
- Three-state outputs
- Typical address access time . . . 35 ns
- TBP28S85A-50 maximum address access time . . . 50 ns
- Typical select time . . . 20 ns
- Typical power . . . 550 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1
9	Q0	21	G4
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

**TBP28S86 ('S478)**

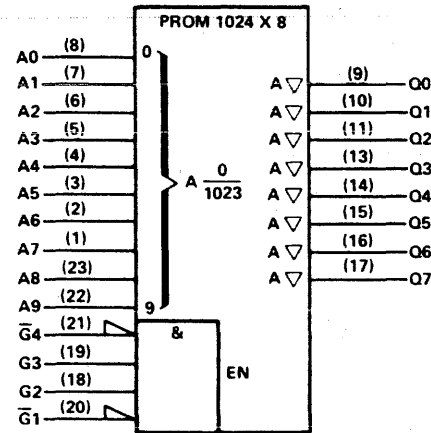
**TBP28S86-60** These products are no longer in production. They are replaced by TBP28S86A and TBP28S86A-50.

**TBP28S86A**  
**TBP28S86A-50**

**PROGRAMMABLE READ-ONLY MEMORIES**

- 1024 X 8
- Three-state outputs
- Typical address access time . . . 35 ns
- TBP28S86A-50 maximum address access time . . . 50 ns
- Typical select time . . . 20 ns
- Typical power . . . 550 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1
9	Q0	21	G4
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

**TBP28SA86 ('S479)**

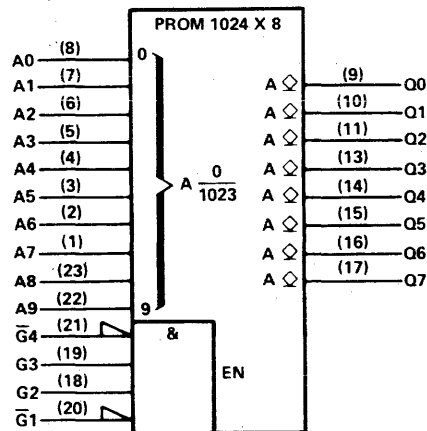
**TBP28SA86-60** These products are no longer in production. They are replaced by TBP28SA86A and TBP28SA86A-50.

† Pin numbers shown on logic symbols are for J and N packages only.  
nc - no internal connection.

**TBP28SA86A**  
**TBP28SA86A-50**  
 PROGRAMMABLE READ-ONLY  
 MEMORIES

- 1024 X 8
- Open-collector outputs
- Typical address access time . . . 35 ns
- TBP28SA86A-50 maximum address access time . . . 50 ns
- Typical select time . . . 20 ns
- Typical power . . . 550 mW

logic symbol†



pin assignments

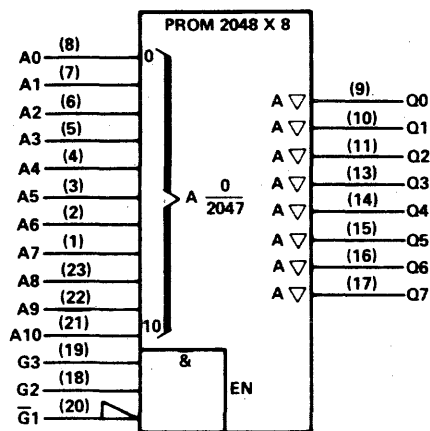
J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1
9	Q0	21	G4
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

**TBP28S165**  
**TBP28S166**  
 PROGRAMMABLE READ-ONLY  
 MEMORIES

- 2048 X 8
- Three-state outputs
- Typical select time . . . 15 ns

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	G2
7	A1	19	G3
8	A0	20	G1
9	Q0	21	A10
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	V <sub>CC</sub>

For chip carrier options and information, contact the factory.

TYPE	PACKAGE ROW SPACING	TYPICAL ADDRESS ACCESS TIME	GUARANTEED MAXIMUM ACCESS TIME	TYPICAL POWER DISSIPATION
TBP28S165	7.62 mm (0.300 in.)	25 ns		550 mW
TBP28S166	15.24 mm (0.600 in.)	35 ns		650 mW

† Pin numbers shown on logic symbols are for J and N packages only.  
 nc — no internal connection.

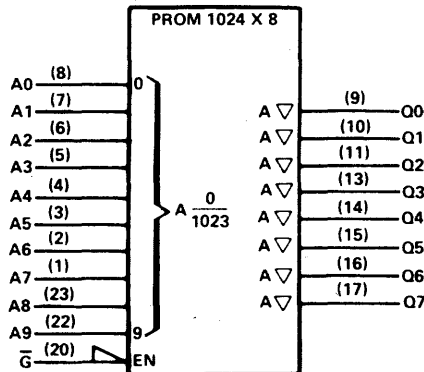
**TBP28S2708 ('S2708)** This product is no longer in production, it is replaced by TBP28S2708A.

**TBP28S2708A**

**PROGRAMMABLE READ-ONLY MEMORIES**

- 1024 X 8
- Three-state outputs
- Typical select time . . . 20 ns
- Typical address access time . . . 35 ns
- Typical power dissipation . . . 550 mW

logic symbol†



pin assignments

J, N PACKAGES			
1	A7	13	Q3
2	A6	14	Q4
3	A5	15	Q5
4	A4	16	Q6
5	A3	17	Q7
6	A2	18	nc
7	A1	19	nc
8	A0	20	G-bar
9	Q0	21	nc
10	Q1	22	A9
11	Q2	23	A8
12	GND	24	VCC

For chip carrier options and information, contact the factory.

**TIM8228** is the same as SN74S428

**TIM8238** is the same as SN74S438

**TIM9905** is the same as SN74LS251

**TIM9906** is the same as SN74LS259

**TIM9907** is the same as SN74148

**TIM9908** is the same as SN74LS348

† Pin numbers shown on logic symbols are for J and N packages only.

nc - no internal connection.



# VT2332/33

## 4096 x 8 STATIC READ ONLY MEMORY

### FEATURES

- 4096 x 8-bit organization
- Single +5 V supply
- Access Time—250 ns (max)
- Totally static operation
- Completely TTL compatible
- VT2332 pin compatible with 2532
- VT2333 pin compatible with 2732
- 3-State Outputs for wired-OR expansion
- Two programmable Chip Selects
- 2708/2716/2532/2732 EPROMs accepted as program data inputs

### DESCRIPTION

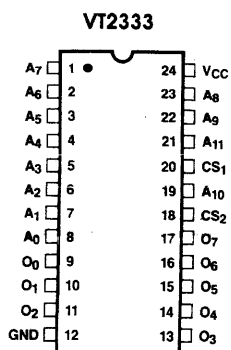
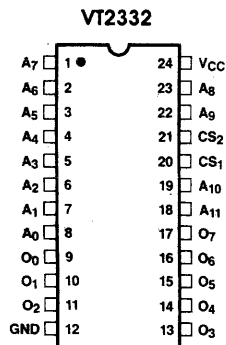
The VT2332/3 high-performance Read Only Memory is organized 4096 words by eight bits with access times of less than 250 ns. This ROM is designed to be compatible with all microprocessor and similar applications where high performance, large bit storage and simple interfacing are important design considerations. This device offers TTL input and output levels with a minimum of 0.4 V noise immunity in conjunction with a +5 V power supply.

The VT2332/3 operates totally asynchronously. No clock input is required. The two programmable

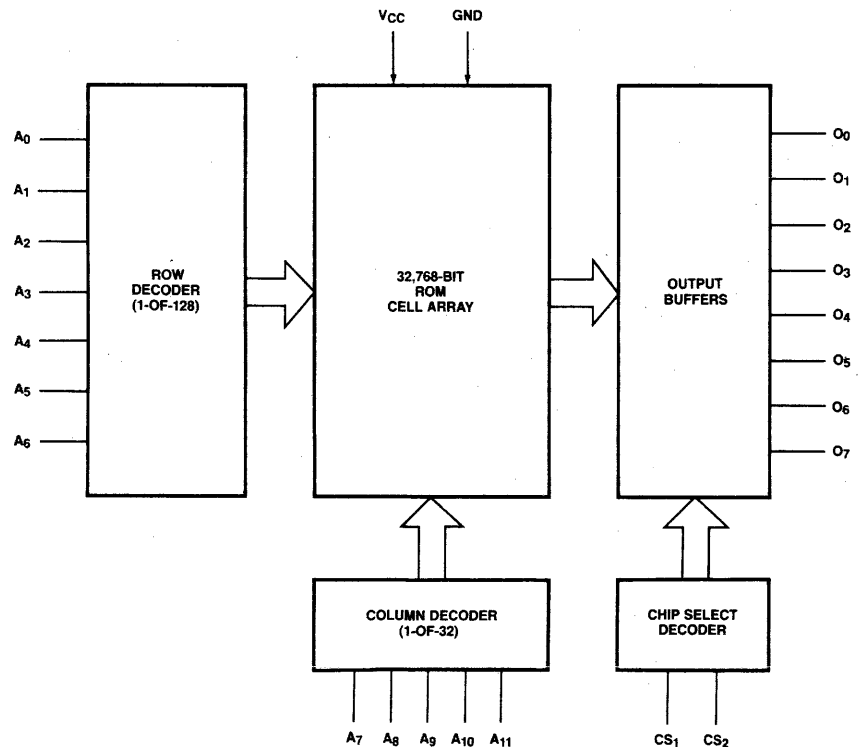
Chip Selects allow up to four 32K ROMs to be wired-OR without external decoding. Both devices offer 3-state output buffers for memory expansion.

Designed to replace either the 2732 or 2532 32K EPROMs, the VT2332/3 can eliminate the need to redesign printed circuit boards for volume mask programmed ROMs after prototyping with EPROMs.

### PIN CONFIGURATIONS



### BLOCK DIAGRAM





# VT2364

## 8192 x 8 STATIC READ ONLY MEMORY

### FEATURES

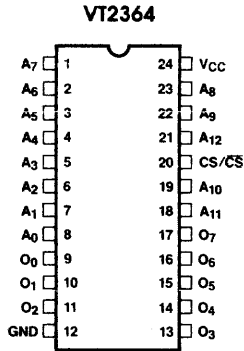
- 2732 EPROM pin compatible
- 8192 x 8-bit organization
- Single + 5 V supply
- Access time—250 ns max
- Totally static operation
- Completely TTL compatible
- Operating Power: 45 mA (max)
- 24-Pin JEDEC approved pin-out
- Programmable Chip Select
- 3-state outputs for wired-OR expansion
- EPROMS accepted as program data input

### DESCRIPTION

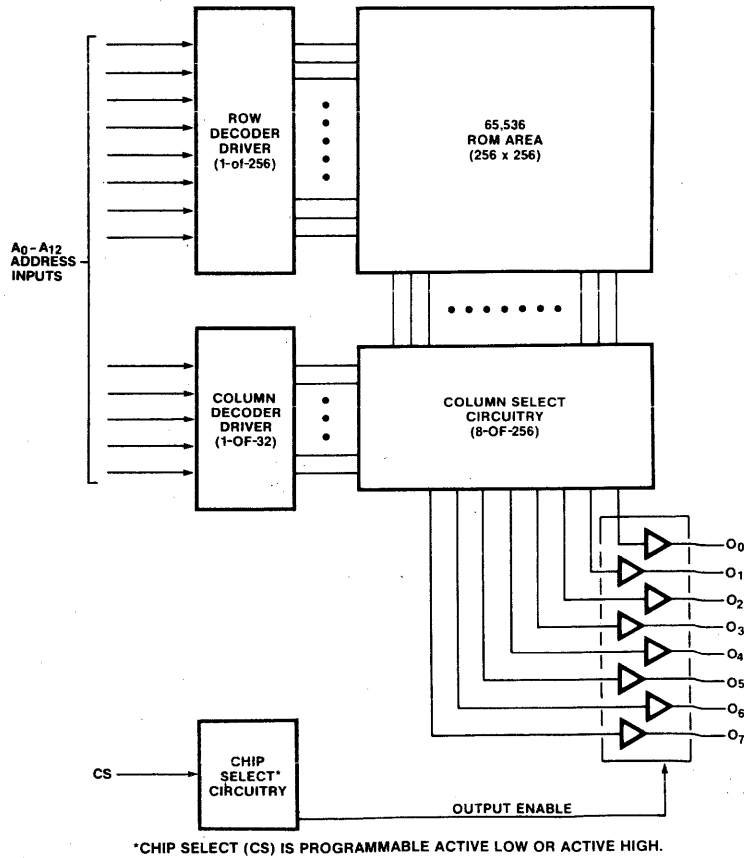
The VT2364 high-performance Read Only Memory is organized as 8,192 words by eight bits with an access time of 250 ns. The ROM is designed to be compatible with all micro-processors and similar applications where high-performance large-bit storage and simple interfacing are important design considerations. It conforms to the JEDEC approved pinouts for 24-pin 64K ROMs.

The VT2364 offers very simple operation with no power down. The programmable Chip Select allows two 64K ROMs to be Wired-OR without external decoding. It is pin compatible with the 2732 EPROM thus eliminating the need to redesign printed circuit boards for volume mask programmed ROMs after prototyping with EPROMS.

### PIN CONFIGURATIONS



### BLOCK DIAGRAM





# VT 2365/66

## 8,192 x 8 STATIC READ ONLY MEMORY

### FEATURES

- 2764/2564 EPROM pin compatible
- 8,192 x 8-bit organization
- Single +5 V supply
- Access time—200 ns (max)
- Totally static operation
- Completely TTL compatible
- Power
  - Operating: 45 mA (max)
  - Standby: 10 mA (max)
- 28-Pin JEDEC approved pinout
- Automatic power down ( $\overline{CE}$ )
- Output Enable function ( $\overline{OE}$ )
- Programmable Chip Select
- 3-state outputs for wired-OR expansion
- EPROMs accepted as program data input

### DESCRIPTION

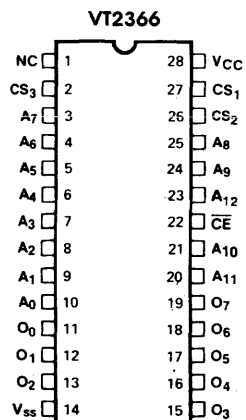
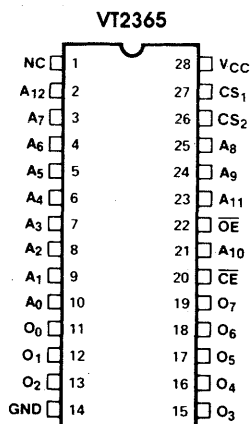
The VT2365/66 high-performance Read Only Memory is organized 8,192 words by eight bits with an access time of 200 ns. The ROM is designed to be compatible with all micro-processors and similar applications where high-performance large-bit storage and simple interfacing are important design considerations. It conforms to the JEDEC approved pinout for 28-pin 64K ROMs.

The VT2365/66 offers an automatic power down feature with power down controlled by the Chip Enable. ( $\overline{CE}$ ) input. When  $\overline{CE}$  goes HIGH, the device will automatically power down and remain in a low-power standby mode as long as  $\overline{CE}$  remains HIGH.

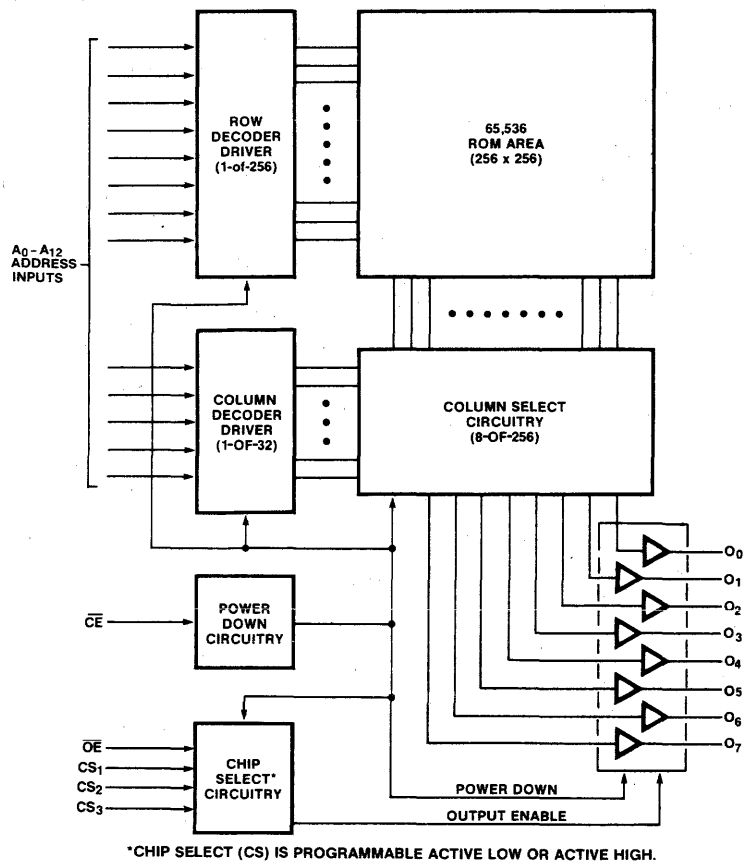
This unique feature provides system level power savings of as much as 90%. The programmable Chip Selects allow up to four 64K ROMs to be wired-OR without external decoding. An additional feature of the VT2365 is the Output Enable ( $\overline{OE}$ ) function which eliminates bus contention in multiple-bus microprocessor systems.

The VT2365 is pin compatible with the 2764 EPROM and the VT2366 is pin compatible with the 2564 EPROM, thus eliminating the need to redesign printed circuit boards for volume mask programmable ROMs after prototyping with EPROMs.

### PIN CONFIGURATIONS



### BLOCK DIAGRAM



\*CHIP SELECT (CS) IS PROGRAMMABLE ACTIVE LOW OR ACTIVE HIGH.

VLSI Technology

MEMORY





# VT23128/29

## 16,384 x 8 STATIC READ ONLY MEMORY

### FEATURES

- 16,384 x 8-bit organization
- Single +5 V supply
- Access time 170 ns max
- Totally static operation
- Completely TTL compatible
- Operating power 45 mA max
- Standby power 10 mA max
- Automatic power down ( $\overline{CE}$ )
- Output enable function ( $\overline{OE}$ )
- Programmable Chip Select
- 3-state outputs for wired-OR expansion
- 28-pin JEDEC approved pinouts
- VT23128 pin compatible with the 2764 EPROM
- VT23129 pin compatible with the 2564 EPROM
- EPROMs accepted as program data input

### DESCRIPTION

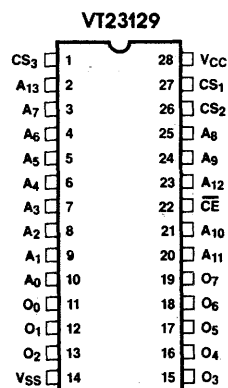
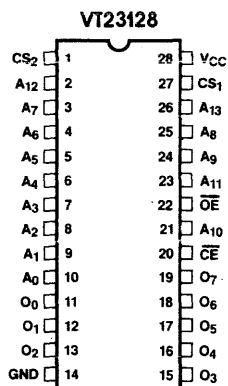
The VT23128/29 high-performance Read Only Memory is organized 16,384 words by eight bits with an access time of 170 ns. It is designed to be compatible with all microprocessors and similar applications where high-performance large-bit storage and simple interfacing are important design considerations.

The VT23128/29 offers automatic power down with power down controlled by the Chip Enable ( $\overline{CE}$ ) input. When  $\overline{CE}$  goes HIGH, the device will automatically power down and remain in a low-power standby mode as long as  $\overline{CE}$  remains HIGH. This unique feature provides system level power savings of as much as 90%. The VT23128 also has an Output Enable ( $\overline{OE}$ ) function to eliminate bus contention in multiple-bus microprocessor systems. The

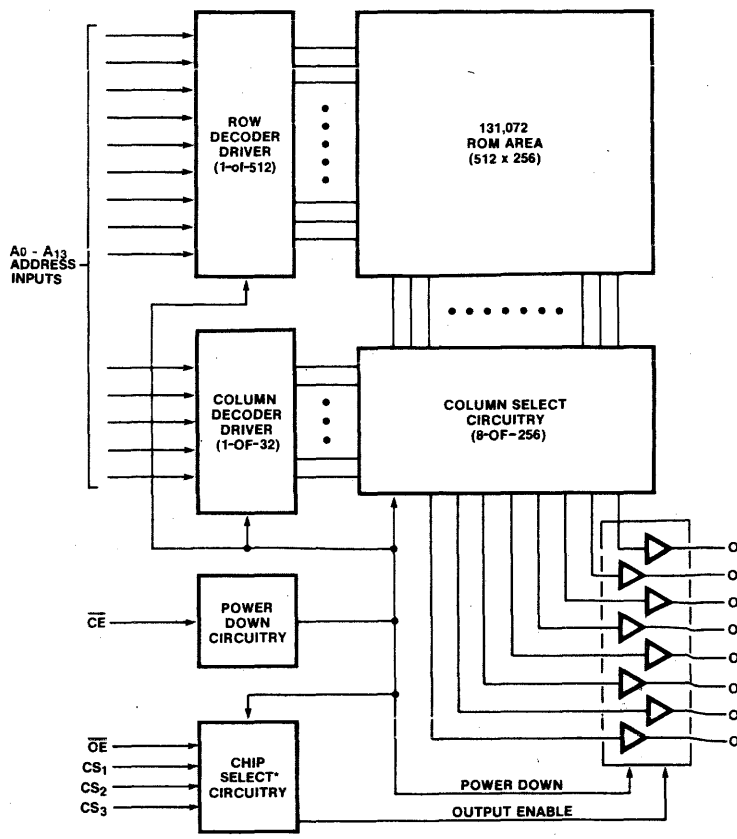
programmable Chip Select allows up to eight 128K ROMs to be wired-OR without external decoding.

The VT23128 is pin compatible with the 2764 EPROM and the VT23129 is pin compatible with the 2564 EPROM, thus eliminating the need to redesign printed circuit boards for volume mask programmable ROMs after prototyping with EPROMs.

### PIN CONFIGURATION



### BLOCK DIAGRAM



\*CHIP SELECT (CS) IS PROGRAMMABLE ACTIVE LOW, ACTIVE HIGH, OR DON'T CARE



# VT23256/57 PRELIMINARY

## 32,768 x 8 STATIC READ ONLY MEMORY

### FEATURES

- 32,768 x 8-bit organization
- Single +5 V supply
- Access time 200 ns max
- Totally static operation
- Completely TTL compatible
- Operating power 60 mA max
- Standby power 10 mA max
- Automatic power down ( $\overline{CE}$ )
- Output Enable function ( $\overline{OE}$ )
- Programmable Chip Select
- 3-state outputs for wired-OR expansion
- 28-pin JEDEC approved pinouts
- VT23256 pin compatible with the 2764 EPROM
- VT23257 pin compatible with the 2564 EPROM
- EPROMs accepted as program data input

### DESCRIPTION

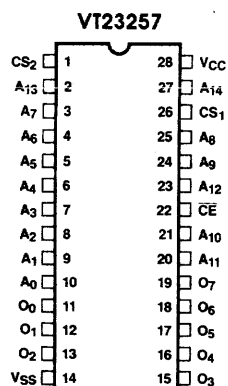
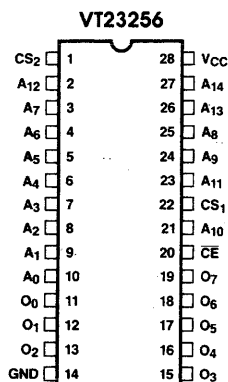
The VT23256/57 high-performance Read Only Memory is organized words by eight bits with an access time of 200 ns. It is designed to be compatible with all microprocessors and similar applications where high-performance large-bit storage and simple interfacing are important design considerations.

The VT23256/57 offers automatic power down with power down controlled by the Chip Enable ( $\overline{CE}$ ) input. When  $\overline{CE}$  goes HIGH, the device will automatically power down and remain in a low power standby mode as long as  $\overline{CE}$  remains HIGH. This unique feature provides system level power savings of as much as 90%. The VT23256/57 also has an Output Enable ( $\overline{OE}$ ) function to eliminate bus contention in

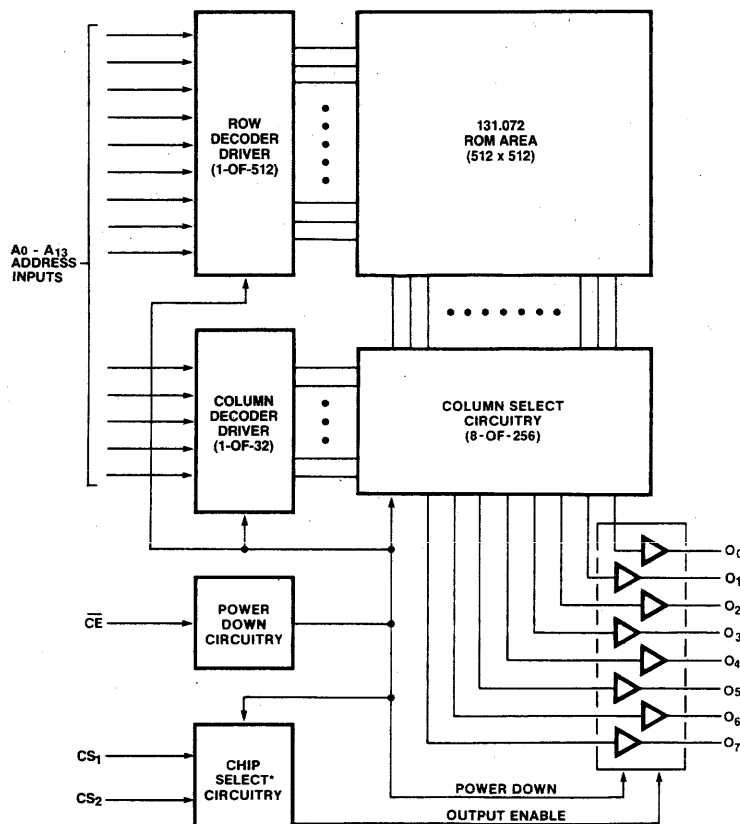
multiple-bus microprocessor systems. The programmable Chip Select allows up to four 256K ROMs to be wired-OR without external decoding.

The VT23256 is pin compatible with the 2764 EPROM and the VT23257 is pin compatible with the 2564 EPROM, thus eliminating the need to redesign printed circuit boards for volume mask programmable ROMs after prototyping with EPROMs.

### PIN CONFIGURATION



### BLOCK DIAGRAM



\*CHIP SELECT (CS) IS PROGRAMMABLE ACTIVE LOW OR ACTIVE HIGH.



# VL4500A

## DYNAMIC RAM CONTROLLER

### FEATURES

- Controls operation of 8K/16K/32K/64K dynamic RAMs
- Creates static RAM appearance
- One package contains address multiplexer, refresh control and timing control
- Directly addresses and drives up to 256K bytes of memory without external drivers
- Operates from microprocessor clock
  - No crystals, delay lines, or RC networks
  - Eliminates arbitration delays
- Refresh may be internally or externally initiated
- Versatile
  - Strap-selected refresh rate
  - Synchronous, predictable refresh
  - Selection of distributed, transparent, and cycle-steal refresh modes
  - Interfaces easily to popular microprocessors

- Strap-selected wait state generation for microprocessor/memory speed matching
- Ability to synchronize or interleave controller with the microprocessor system (including multiple controllers)
- Three-state outputs allow multiport memory configuration
- Performance ranges of 150 ns/200 ns/250 ns
- Compatible with TI TMS 4500A

### DESCRIPTION

The VL4500A is a monolithic DRAM system controller designed to provide address multiplexing, timing, control and refresh/access arbitration functions to simplify the interface of dynamic RAMs to microprocessor systems.

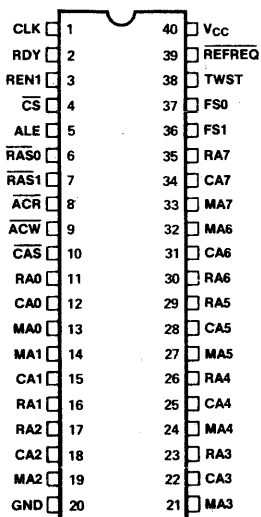
The controller contains a 16-bit multiplexer that generates the address lines for the memory device from the 16 system address bits and provides the strobe signals required by the memory to decode the address. An 8-bit refresh counter generates the 256-row addresses required to refresh.

A refresh timer is provided that generates the necessary timing to refresh the dynamic memories and assure data retention.

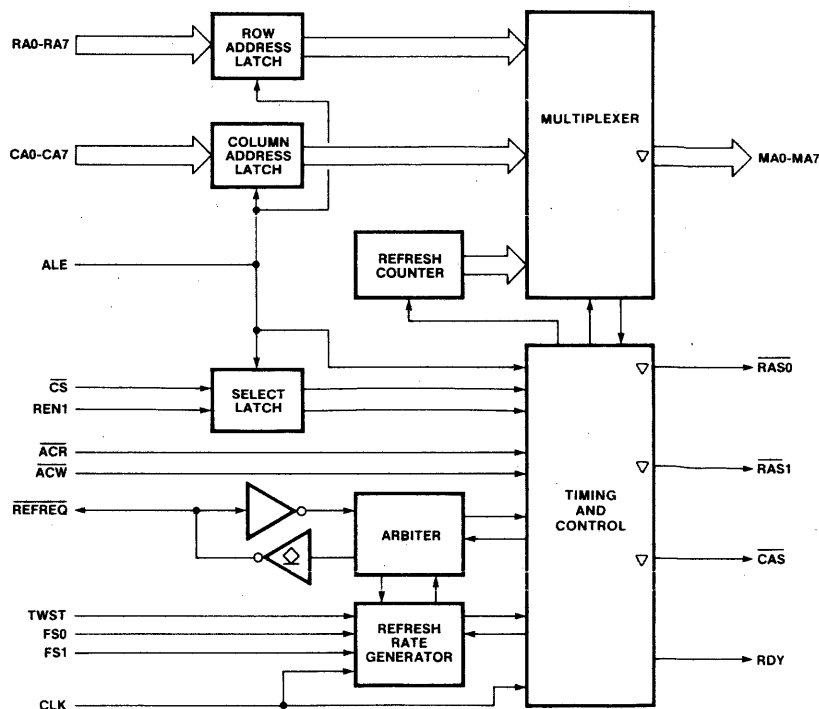
The VL4500A also contains refresh/access arbitration circuitry to resolve conflicts between memory access requests and memory refresh cycles. The VL4500A is offered in a 40-pin, 600-mil dual in-line plastic package and is guaranteed for operation from 0°C to 70°C.

### PIN CONFIGURATION

(Top View)



### BLOCK DIAGRAM



**ABSOLUTE MAXIMUM RATINGS**

Operating Ambient Temperature Range	0°C to 70°C
Storage Temperature Range	-65°C to 150°C
Supply Voltage Range, $V_{CC}^{(1)}$	-1.5 to 7 V
Input Voltage Range (any input) <sup>(1)</sup>	-1.5 to 7 V
Continuous Power Dissipation	1.2 W

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of this device at these or any other conditions above those indicated on the

operational sections of this specification is not implied and exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**RECOMMENDED OPERATING CONDITIONS:  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$** 

Symbol	Parameter	Min	Nom	Max	Unit
$V_{CC}$	Supply Voltage	4.5	5.0	5.5	V
$V_{IH}$ (except $\overline{\text{REFREQ}}$ )	High-Level Input Voltage	2.0		6.0	V
$V_{IH}$ ( $\overline{\text{REFREQ}}$ )	High-Level Input Voltage	2.4		6.0	V
$V_{IL}$ (except $\overline{\text{REFREQ}}$ )	Low-Level Input Voltage	-1.0 <sup>(2)</sup>		0.8	V
$V_{IL}$ ( $\overline{\text{REFREQ}}$ )	Low-Level Input Voltage	-1.0 <sup>(2)</sup>		1.2	V

**DC CHARACTERISTICS:  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5\text{ V} \pm 10\%$** 

Symbol	Parameter		Test Conditions	Min	Typ <sup>(3)</sup>	Max	Unit
$V_{OH}$	High-Level Output Voltage	MA0-MA7, RDY	$I_{OH} = -1\text{ mA}$ $V_{CC} = 4.5\text{ V}$	2.4			V
		RAS0, RAS1, CAS		2.7			
		$\overline{\text{REFREQ}}$		2.4			
$V_{OL}$	Low-Level Output Voltage		$I_{OL} = 4\text{ mA}$ $V_{CC} = 4.5\text{ V}$			0.4	V
$I_{IH}$	High-Level Input Current	$\overline{\text{REFREQ}}$	$V_I = 5.5\text{ V}$			100	$\mu\text{A}$
		All others				10	
$I_{IL}$	Low-Level Input Current	$\overline{\text{REFREQ}}$	$V_I = 0\text{ V}$			-1.25	mA
		All others				-10	$\mu\text{A}$
$I_{OZ}$	Off-State Output Current		$V_O = 0$ to $4.5\text{ V}$ $V_{CC} = 5.5\text{ V}$			$\pm 50$	$\mu\text{A}$
$I_{CC}$	Operating Supply Current		$T_A = 0^\circ\text{C}$ $V_{CC} = 5.5\text{ V}$		100	140	mA
$C_i$	Input Capacitance		$V_I = 0\text{ V}$ $f = 1\text{ MHz}$		5		pF
$C_o$	Output Capacitance		$V_O = 0\text{ V}$ $f = 1\text{ MHz}$		6		pF

**Notes:**

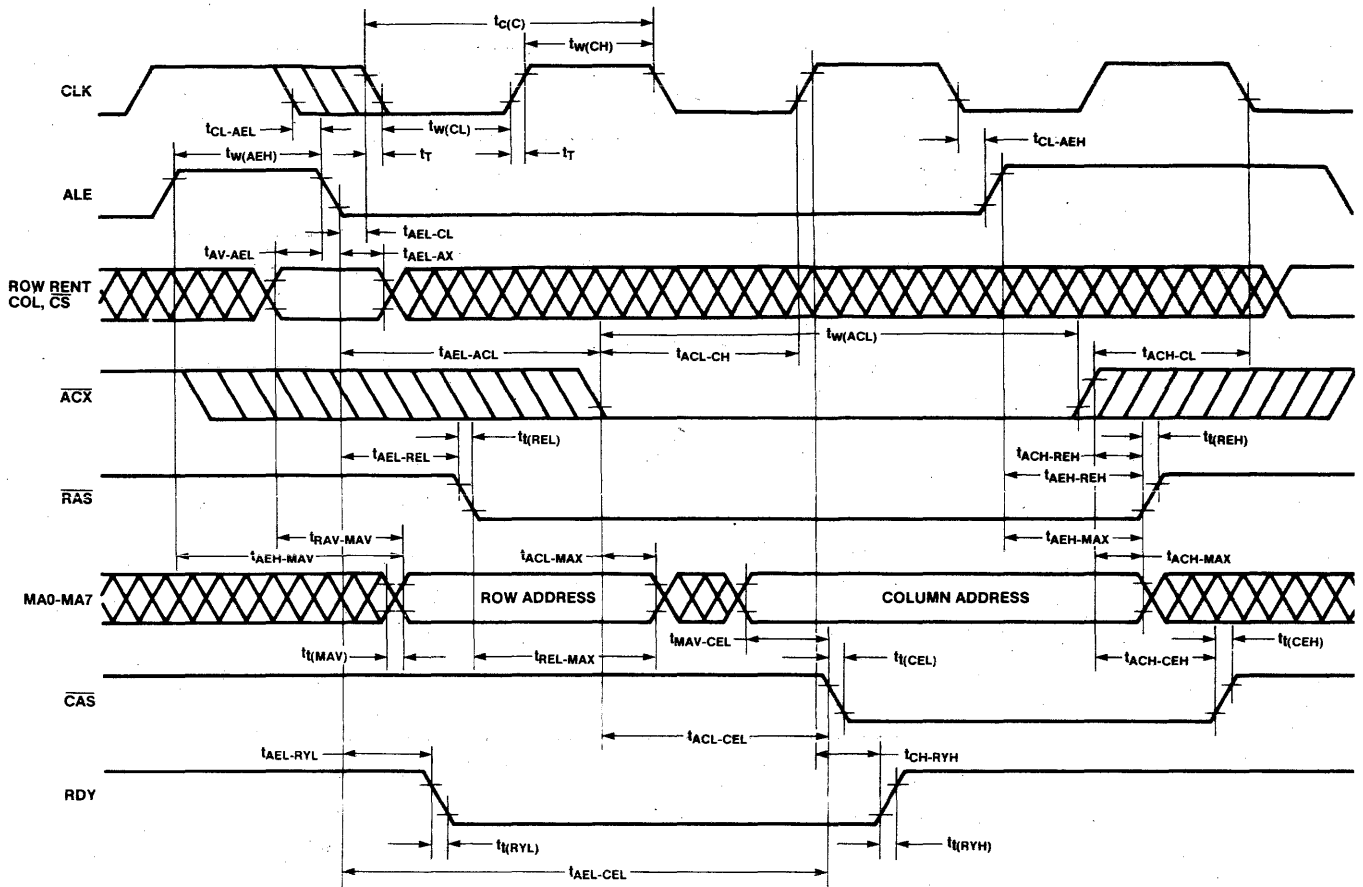
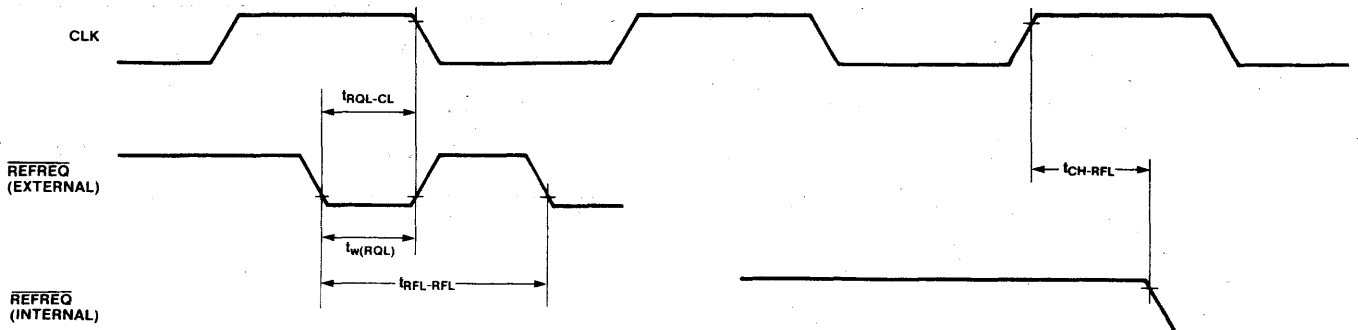
1. Voltage values are with respect to the ground terminal.
2. The algebraic convention, where the more negative limit is designated as minimum, is used in this data sheet for logic voltage levels only.
3. All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  except where otherwise noted.

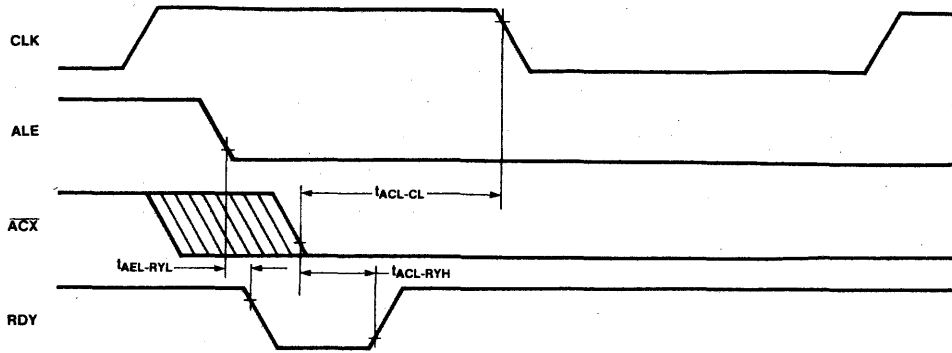
**AC CHARACTERISTICS:  $T_A = 0^\circ\text{C}$  to  $+70^\circ\text{C}$ ,  $V_{CC} = 5\text{V} \pm 10\%$** 

Symbol	Parameter	VL4500A-15PC		VL4500A-20PC		VL4500A-25PC		Unit
		Min	Max	Min	Max	Min	Max	
$t_{C(C)}$	CLK Cycle Time	100		100		140		ns
$t_{W(CH)}$	CLK High Pulse Width	20		20		20		
$t_{W(CL)}$	CLK Low Pulse Width	35		35		35		
$t_t$	Transition Time, All Inputs		50		50		50	
$t_{AEL-CL}$	Time Delay, ALE Low to CLK Starting Low(1)	10		10		15		
$t_{CL-AEL}$	Time Delay, CLK Low to ALE Starting Low(1)	10		10		15		
$t_{CL-AEH}$	Time Delay, CLK Low to ALE Starting High(2)	15		20		20		
$t_{W(AEH)}$	Pulse Width ALE High	50		60		60		
$t_{AV-AEL}$	Time Delay, Address, REN1, CS Valid to ALE Low	5		10		15		
$t_{AEL-AX}$	Time Delay, ALE Low to Address Not Valid	10		10		10		
$t_{AEL-ACL}$	Time Delay, ALE Low to ACX Low(3, 4, 5, 6)	$t_{h(RA)} + 30$		$t_{h(RA)} + 40$		$t_{h(RA)} + 50$		
$t_{ACH-CL}$	Time Delay, ACX High to CLK Low(3, 7)	20		20		20		
$t_{ACL-CH}$	Time Delay, ACX Low to CLK Starting High (to remove RDY)	30		30		30		
$t_{RQL-CL}$	Time Delay, REFREQ Low to CLK Starting Low(8)	20		20		20		
$t_{W(RQL)}$	Pulse Width, REFREQ Low	20		20		20		

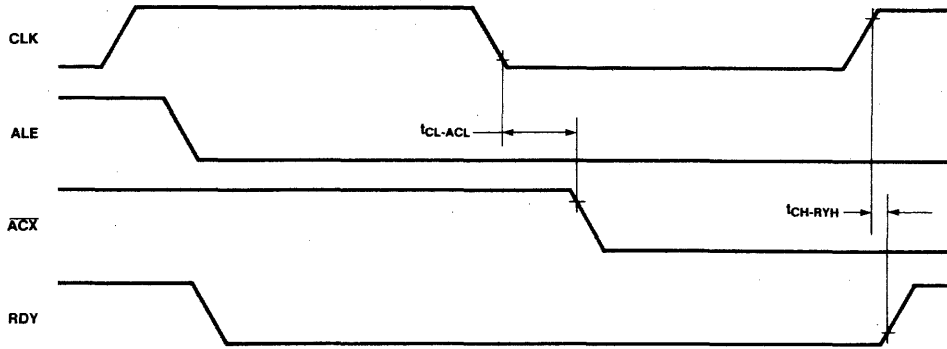
**Notes:**

1. Coincidence of the trailing edge of CLK and the trailing edge of ALE should be avoided, as the refresh/access occurs on the trailing CLK edge. A trailing edge of CLK should occur during the interval from ACX high to ALE low.
2. If ALE rises before ACX and a refresh request is present, the falling edge of CLK after  $t_{CL-AEH}$  will output the refresh address to MA0-MA7 and initiate a refresh cycle.
3. These specifications relate to system timing and do not directly reflect device performance.
4. On the access grant cycle following refresh, the occurrence of CAS low depends on the relative occurrence of ALE low to ACX low. If ACX occurs prior to or coincident with ALE then CAS is timed from the CLK high transition that causes RAS low. If ACX occurs 20 ns or more after ALE then CAS is timed from the CLK low transition following the CLK high transition causing RAS low.
5. For maximum speed access (internal delays on both access and access grant cycles), ACX should occur prior to or coincident with ALE.
6.  $t_{h(RA)}$  is the dynamic memory row address hold time. ACX should follow ALE by  $t_{AEL-CEL}$  in systems where the required  $t_{h(RA)}$  is greater than  $t_{REL-MAX}$  minimum.
7. Minimum of 20 ns is specified to ensure arbitration will occur on falling CLK edge.  $t_{ACH-CL}$  also affects precharge time such that the minimum  $t_{ACH-CL}$  should be equal or greater than:  $t_{W(RH)} - t_{W(CL)} + 30$  ns (for cycle where ACX high occurs prior to ALE high) where  $t_{W(RH)}$  is the DRAM RAS precharge time.
8. This parameter is necessary only if refresh arbitration is to occur on this low-going CLK edge (in systems where refresh is synchronized to external events).

**TIMING DIAGRAMS**  
**ACCESS CYCLE TIMING**

**REFRESH REQUEST TIMING**


**READY (RDY) SIGNAL TIMING (WAIT STATE OPERATION, TWST = 1)**


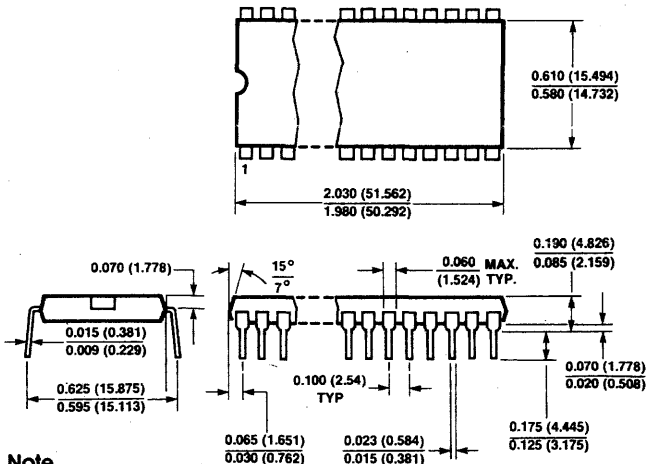
RDY starting high is timed from  $\overline{ACX}$  low ( $t_{ACL-RYH}$ ) for the condition  $\overline{ACX}$  going low while CLK high.



RDY starting high is timed from CLK high ( $t_{CH-RYH}$ ) for the condition  $\overline{ACX}$  going low while CLK low.

**PACKAGE DIAGRAM**

40-LEAD PLASTIC DUAL IN-LINE  
(P-DIP) PACKAGE


**Note**

1. All metric dimensions are in parenthesis ( ).

**ORDERING INFORMATION**

VL4500A-15PC (TMS4500AN-15)	150 ns	Plastic, commercial, 0 - 70°C
VL4500A-20PC (TMS4500AN-20)	200 ns	Plastic, commercial, 0 - 70°C
VL4500A-25PC (TMS4500AN-25)	250 ns	Plastic, commercial, 0 - 70°C

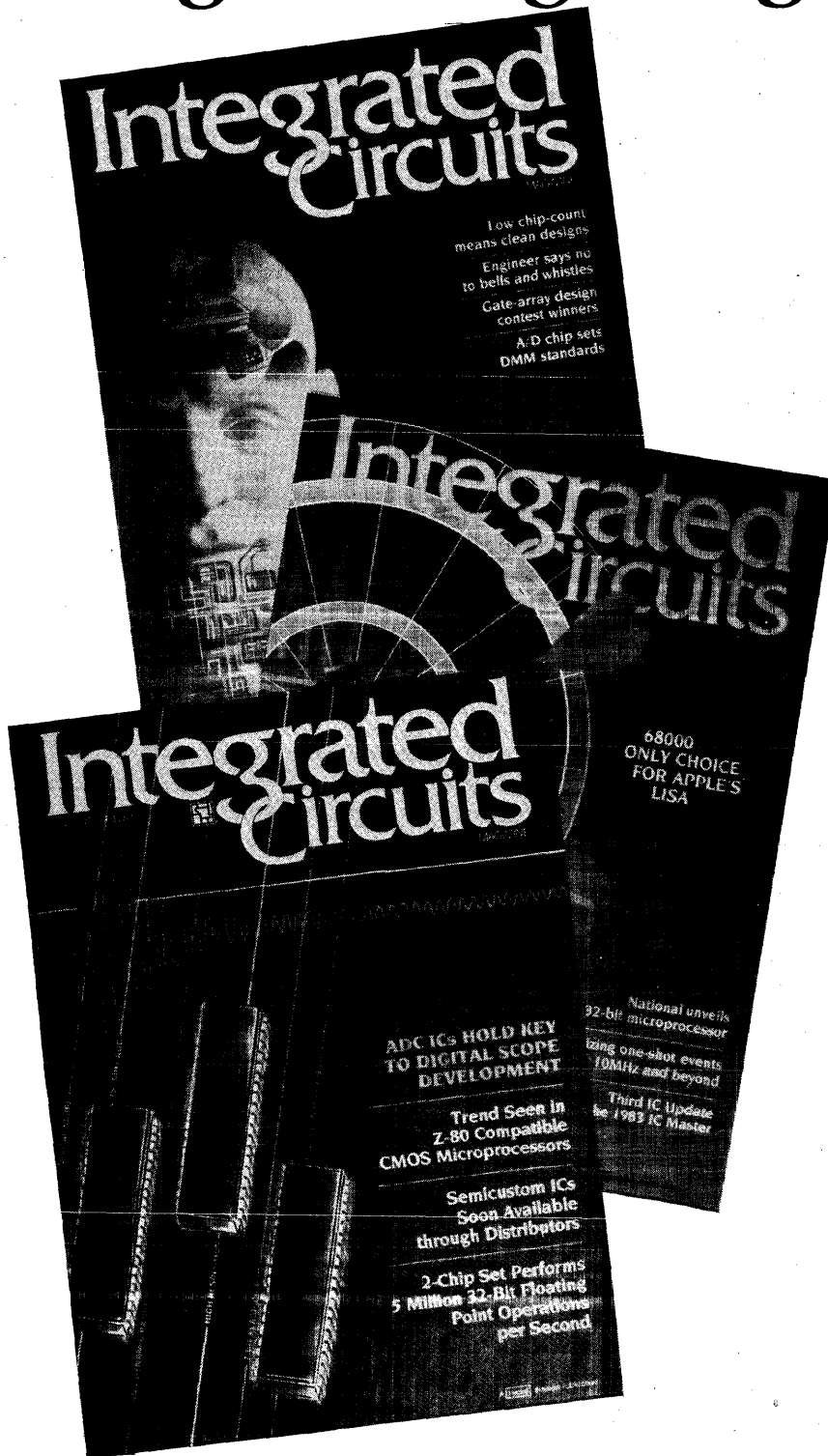
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# NEW PRODUCT BULLETIN

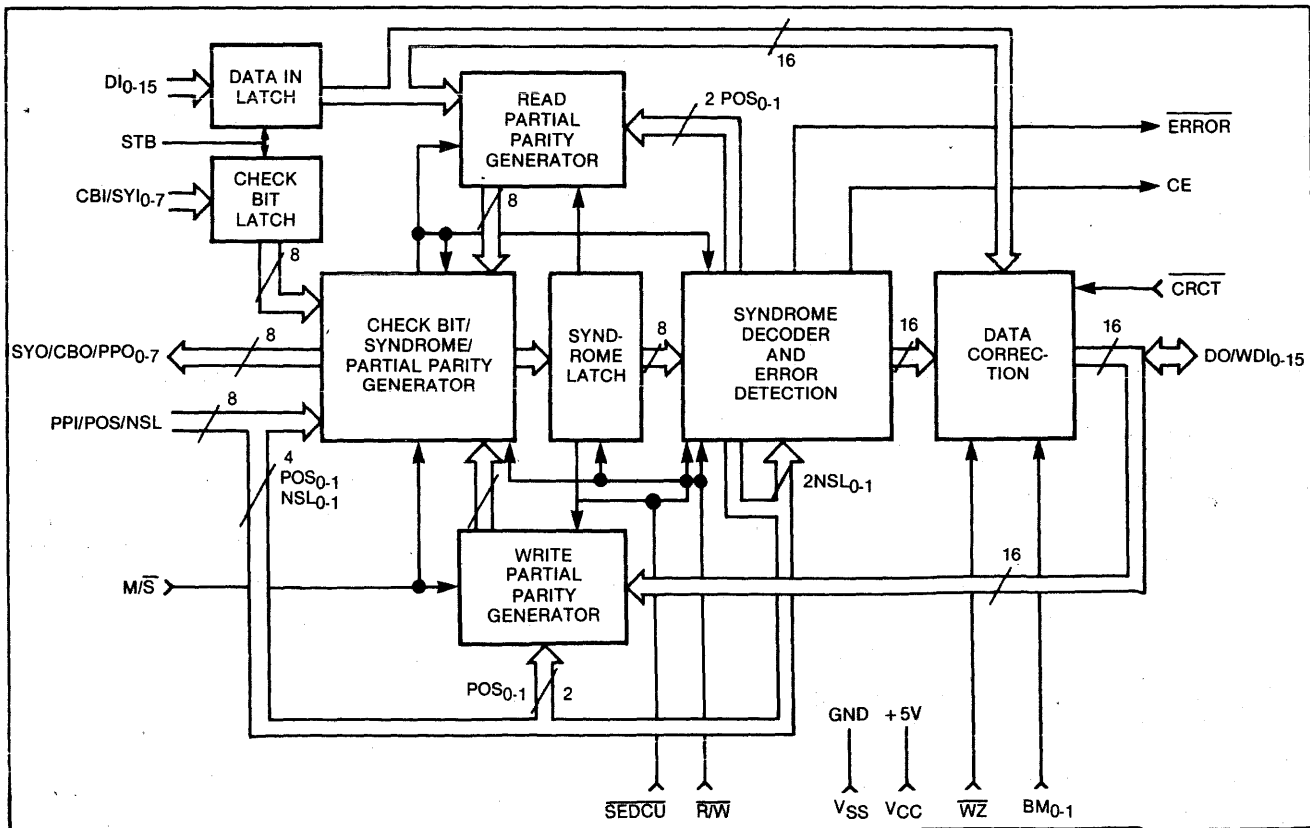
## Error Detection and Correction Unit WD8206

### FEATURES

- DETECTS AND CORRECTS ALL SINGLE BIT ERRORS
- DETECTS ALL DOUBLE BIT AND MOST MULTIPLE BIT ERRORS
- 52 NS MAXIMUM FOR DETECTION; 67 NS MAXIMUM FOR CORRECTION (16 BIT SYSTEM)
- EXPANDABLE TO HANDLE 80 BIT MEMORIES
- SYNDROME OUTPUTS FOR ERROR LOGGING
- SEPARATE INPUT AND OUTPUT BUSES — NO TIMING STROBES REQUIRED
- SUPPORTS READS WITH AND WITHOUT CORRECTION, WRITES, PARTIAL (BYTE) WRITES, AND READ-MODIFY-WRITES
- NMOS TECHNOLOGY FOR LOW POWER
- COMPATIBLE WITH INTEL 8206 DEVICE
- 68 PIN LEADLESS JEDEC PACKAGE
- SINGLE +5V SUPPLY

### DESCRIPTION

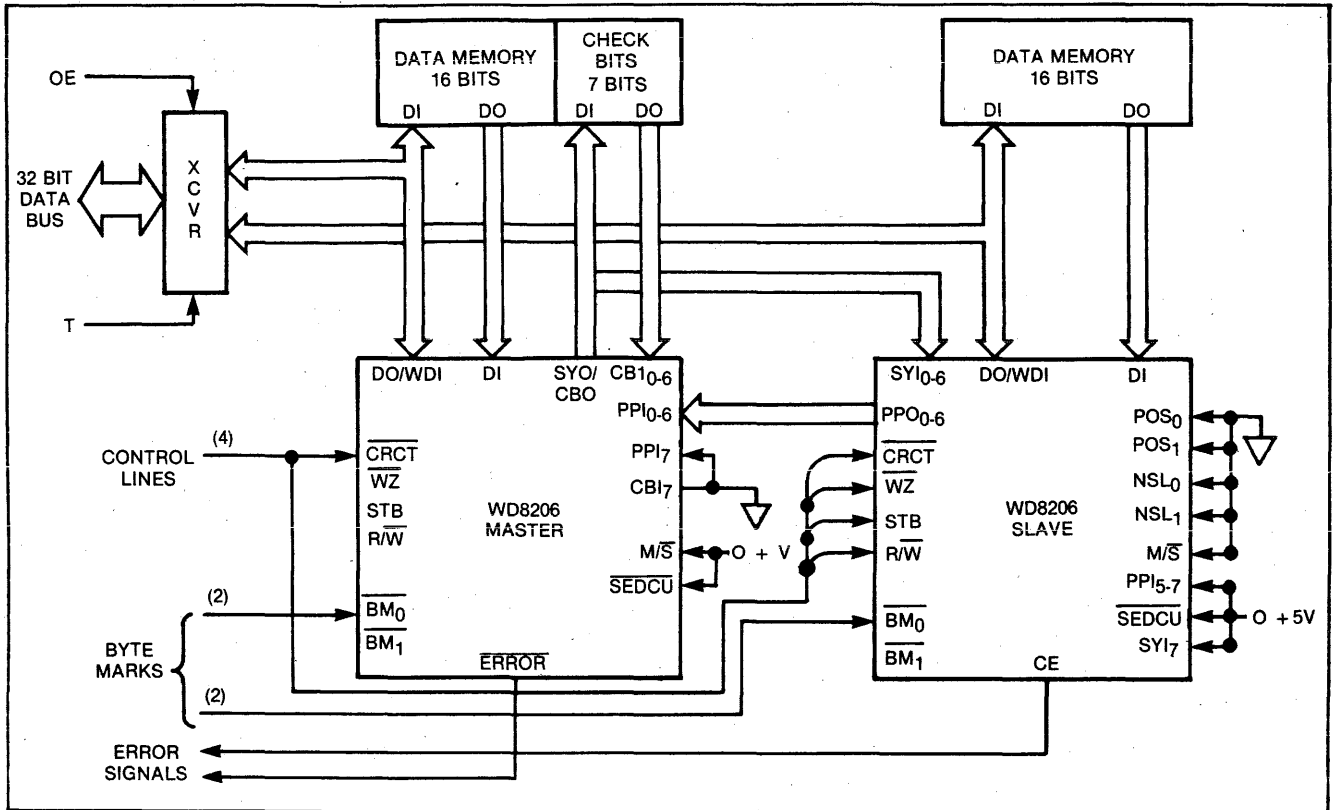
The NMOS 8206 Error Detection and Correction Unit is a high-speed device that provides error detection and correction for memory systems (static and dynamic) requiring high reliability and performance. Each WD8206 handles 8 or 16 data bits and up to 8 check bits. WD8206's can be cascaded to provide correction and detection for up to 80 bits of data. Other WD8206 features include the ability to handle byte writes, memory initialization, and error logging.



**WD8206 BLOCK DIAGRAM**

## PROPOSED APPLICATIONS

The WD8206 interface to a typical 32 bit memory system is illustrated below. For larger systems, the partial parity bits from slaves two to four must be XOR'd externally, which calls for one level of XOR gating for three WD8206's and two levels for four or five WD8206's.



SIMPLIFIED SYSTEM DIAGRAM

# NEW PRODUCT BULLETIN

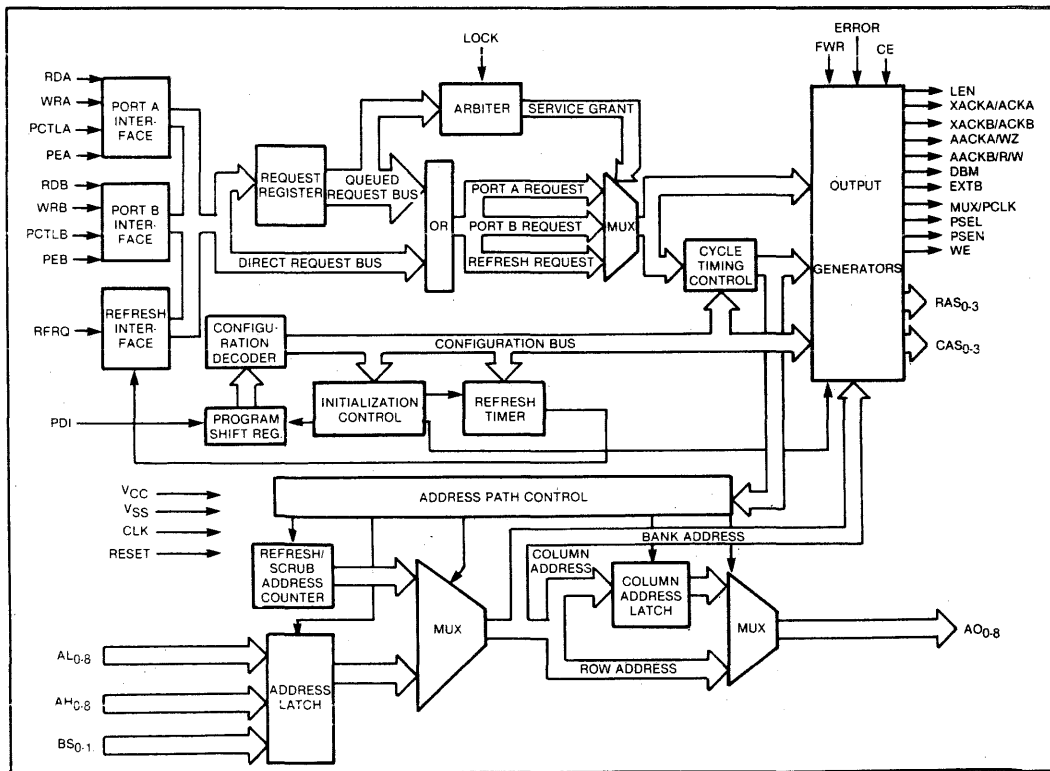
## Advanced Dynamic RAM Controller WD8207

### FEATURES

- PROVIDES ALL SIGNALS NECESSARY TO CONTROL 16K, 64K AND 256K DYNAMIC RAMS
- DIRECTLY ADDRESSES AND DRIVES UP TO 2 MEGABYTES WITHOUT EXTERNAL DRIVERS
- SUPPORTS SINGLE AND DUAL-PORT CONFIGURATIONS
- AUTOMATIC RAM INITIALIZATION IN ALL MODES
- FIVE PROGRAMMABLE REFRESH MODES
- TRANSPARENT MEMORY SCRUBBING IN ECC MODE
- DATA TRANSFER ACKNOWLEDGE SIGNALS FOR EACH PORT
- PROVIDES SIGNALS TO DIRECTLY CONTROL THE WD8206 ERROR DETECTION AND CORRECTION UNIT
- SUPPORTS SYNCHRONOUS OR ASYNCHRONOUS OPERATION ON EITHER PORT
- SINGLE +5V SUPPLY

### DESCRIPTION

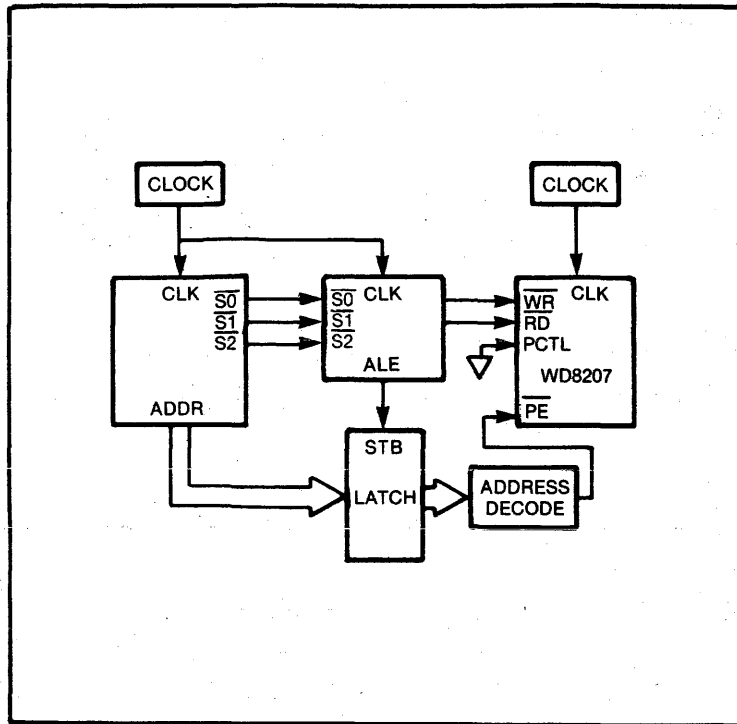
The WD8207 is a 68-pin leadless JEDEC type A hermetic chip carrier. The WD8207 Advanced Dynamic RAM Controller (ADRC) is a high-performance, systems-oriented, Dynamic RAM controller that is designed to easily interface 16K, 64K and 256K Dynamic RAMs microprocessor Systems. A dual-port interface allows two different busses to independently access memory. When configured with a WD8206 Error Detection and Correction Unit the WD8207 supplies the necessary logic for designing large error-corrected memory arrays. This combination provides automatic memory initialization and transparent memory error scrubbing.



**WD8207 BLOCK DIAGRAM**

## PROPOSED APPLICATIONS

The ADRC supports several microprocessor interface options including synchronous and asynchronous connection.



**SIMPLIFIED SYSTEM DIAGRAM**

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**X2444**  
**X2444I**

**256 Bit Serial**  
**Nonvolatile Static RAM**

**16x16 Bit Serial NOV RAM\***

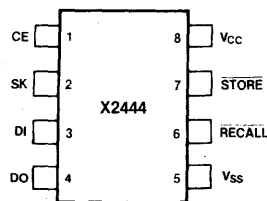
- Ideal for use with Single Chip Microcomputers
  - Static Timing 1 MHz Operation
  - Minimum I/O Interface
  - Serial Port Compatible (COPST™, 8051)
  - Minimum Support Circuits
- Software and Hardware Control of Nonvolatile Functions
  - Maximum Store Protection
- TTL Compatible
- 16x16 Organization
- X2444 Emulator Available—X2444E
- Reliable N-Channel Floating Gate Technology
- Single 5V Supply
- Low Power Dissipation
  - Active Current: 15mA Typ.
  - Store Current: 8mA Typ.
  - Standby Current: 6mA Typ.
  - Sleep Current: 5mA Typ.
- 8-Pin Mini-Dip
  - Low Cost
  - Compact

The X2444 is a 256-bit Xicor NOV RAM memory fabricated in Xicor's proven NMOS technology. The device is organized as 16 words of 16 bits each. Serial access allows the use of a cost-effective 8-pin package, making the X2444 ideal for cost sensitive and compact design applications. Each bit of the static RAM is overlaid with a bit of nonvolatile electrically erasable PROM (E<sup>2</sup>PROM). Data can be transferred back and forth between the two memories either by instructions sent from the processor over the serial interface, or by toggling the external STORE and RECALL inputs. Nonvolatile data is retained in the E<sup>2</sup>PROM, while independent data can be accessed and updated in the RAM.

High voltage pulses and supplies are not required. A single 5V supply is the only power source needed, and all signals are TTL compatible.

The X2444 offers many modes of operation in order to minimize the power consumption of the chip. The chip is placed in the STANDBY mode whenever it is deselected, and is placed in the SLEEP mode whenever the sleep instruction is executed. The chip will automatically return active from STANDBY when selected by CE, and will exit the SLEEP mode when the next RECALL operation is performed, either by the RCL instruction or by taking the RECALL input low.

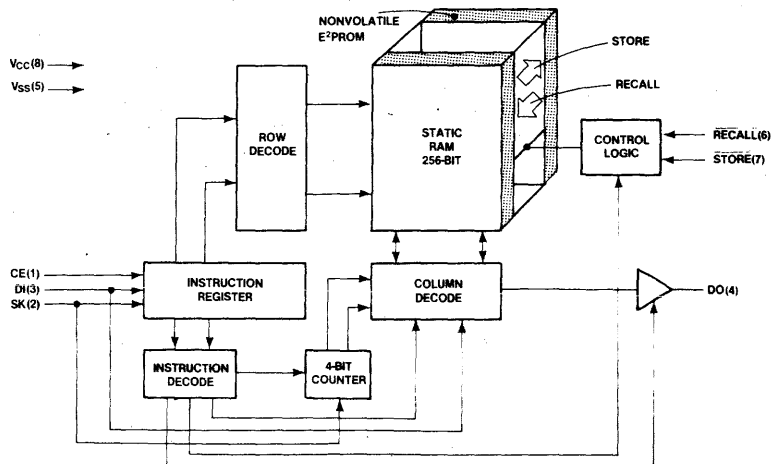
**PIN CONFIGURATION**  
**8-PIN DIP .300"**



**PIN NAMES**

CE	CHIP ENABLE
SK	SERIAL CLOCK
DI	SERIAL DATA IN
DO	SERIAL DATA OUT
RECALL	RECALL
STORE	STORE
V <sub>CC</sub>	+ 5V
V <sub>SS</sub>	GROUND

**FUNCTIONAL DIAGRAM X2444 (16x16)**



\*NOV RAM is Xicor's nonvolatile static random access memory device.  
COPST™ is a trademark of National Semiconductor, Inc.

©Xicor, 1983 Patents Pending  
Characteristics Subject to Change  
Without Notice  
June 1983, Stock No. 200-019

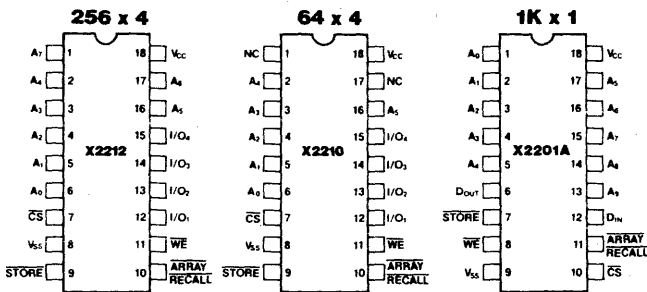


**Xicor NOVRAM<sup>\*</sup> Memories**  
**Nonvolatile Static RAMS**

5V-Programmable	X2212	256 x 4 Bit Nonvolatile Static RAM
10% Power Supply Margin	X2210	64 x 4 Bit Nonvolatile Static RAM
V <sub>CC</sub> = 5V ± 10%	X2201A	1024 x 1 Bit Nonvolatile Static RAM

- **NONVOLATILE STATIC RAM:** The X2201A, X2210, and X2212 are organized as conventional static RAMs overlaid bit-for-bit with a nonvolatile Electrically Erasable PROM (E<sup>2</sup>PROM). Nonvolatile data can be stored in the E<sup>2</sup>PROM and at the same time independent data can be accessed in the RAM memory. At any time, data can be transferred back-and-forth between the RAM and E<sup>2</sup>PROM by simple store and array recall signals.
- **5V ONLY:** High-voltage pulses or supplies are never required. A single 5V supply is the *only power source* ever required for any function.
- **EASE-OF-USE:** Unprecedented simplicity, all inputs and outputs are directly TTL compatible. Fully static timing. Three-state output. 18-pin package.
- **PERFORMANCE:** RAM cycle time is less than 300ns. During the lifetime of the device, data can be recalled from the E<sup>2</sup>PROM an unlimited number of times.
- **POWER-FAILURE PROTECTION:** One simple TTL signal saves the entire RAM database. A snapshot nonvolatile copy of all RAM data is internally stored safely without power and can be recalled to the RAM when power returns. No battery backup required.
- **ORGANIZED FOR MICROCOMPUTER SYSTEMS:** The common data input and output is organized four bits wide on the X2210 and X2212. The X2201A is organized conveniently by one for larger memory applications.
- **Xicor's products are fabricated with reliable n-channel floating gate MOS technology.** For systems where RAM nonvolatility or *in-the-circuit* ROM changes by TTL signals are important, the Xicor X2201A, X2210, or X2212 is the ideal choice.

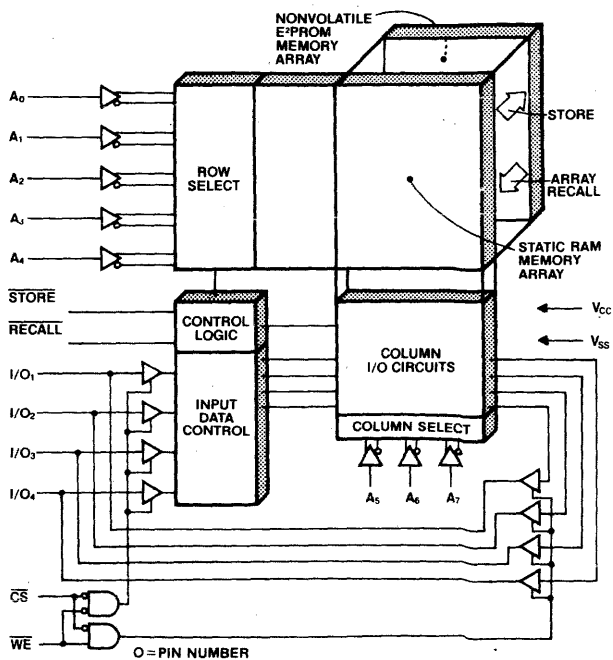
**PIN CONFIGURATIONS**



**PIN NAMES**

A <sub>0</sub> —A <sub>9</sub>	ADDRESS INPUTS
I/O <sub>1</sub> —I/O <sub>4</sub>	DATA INPUT/OUTPUT
WE	WRITE ENABLE
CS	CHIP SELECT
ARRAY RECALL	ARRAY RECALL
STORE	STORE
V <sub>CC</sub>	+5V
V <sub>SS</sub>	GROUND
NC	NO CONNECT

**FUNCTIONAL DIAGRAM**



© XICOR, 1983 Patents pending.

July 1983 Stock No. 200-024

\*NOVRAM is Xicor's nonvolatile static RAM device

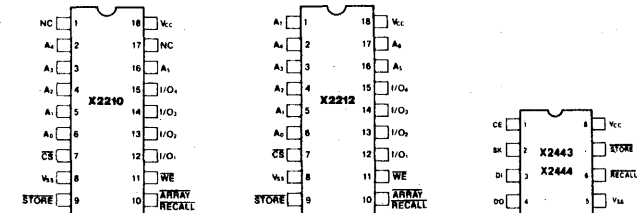
## XICOR ORDERING INFORMATION

Device Type	Part #	Organization	Package	Temp. Range	Processing	Access Time	
NOVRAM	X2444P	16×16*	Plastic	0° to +70°C	Standard	N/A	
	X2444PI	16×16*	Plastic	-40° to +85°C	Standard	N/A	
	X2443P	16×16*	Plastic	0° to +70°C	Standard	N/A	
	X2443PI	16×16*	Plastic	-40° to +85°C	Standard	N/A	
	X2210D	64×4	Cerdip	0° to +70°C	Standard	300ns	
	X2210DI	64×4	Cerdip	-40° to +85°C	Standard	300ns	
	X2210DM	64×4	Cerdip	-55° to +125°C	Standard	300ns	
	X2210DMB	64×4	Cerdip	-55° to +125°C	883B	300ns	
	X2212D	256×4	Cerdip	0° to +70°C	Standard	300ns	
	X2212DI	256×4	Cerdip	-40° to +85°C	Standard	300ns	
	X2212DM	256×4	Cerdip	-55° to +125°C	Standard	300ns	
	X2212DMB	256×4	Cerdip	-55° to +125°C	883B	300ns	
	E <sup>2</sup> PROM	X2804AD	512×8	Cerdip	0° to +70°C	Standard	300ns
		X2804AD-35	512×8	Cerdip	0° to +70°C	Standard	350ns
X2804AD-45		512×8	Cerdip	0° to +70°C	Standard	450ns	
X2804ADI-35		512×8	Cerdip	-40° to +85°C	Standard	350ns	
X2804ADI-45		512×8	Cerdip	-40° to +85°C	Standard	450ns	
X2804ADM-35		512×8	Cerdip	-55° to +125°C	Standard	350ns	
X2804ADM-45		512×8	Cerdip	-55° to +125°C	Standard	450ns	
X2804ADM-45		512×8	Cerdip	-55° to +125°C	883B	350ns	
X2804ADM-45		512×8	Cerdip	-55° to +125°C	883B	450ns	
X2816AD		2K×8	Cerdip	0° to +70°C	Standard	300ns	
X2816AD-35		2K×8	Cerdip	0° to +70°C	Standard	350ns	
X2816AD-45		2K×8	Cerdip	0° to +70°C	Standard	450ns	
X2816ADI-35		2K×8	Cerdip	-40° to +85°C	Standard	350ns	
X2816ADI-45		2K×8	Cerdip	-40° to +85°C	Standard	450ns	
X2816ADM-35		2K×8	Cerdip	-55° to +125°C	Standard	350ns	
X2816ADM-45		2K×8	Cerdip	-55° to +125°C	Standard	450ns	
X2816ADM-45		2K×8	Cerdip	-55° to +125°C	883B	350ns	
X2816ADM-45		2K×8	Cerdip	-55° to +125°C	883B	450ns	
X2816AE		2K×8	LCC	0° to +70°C	Standard	300ns	
X2816AE-35		2K×8	LCC	0° to +70°C	Standard	350ns	
X2816AE-45		2K×8	LCC	0° to +70°C	Standard	450ns	
X2816AEI-35		2K×8	LCC	-40° to +85°C	Standard	350ns	
X2816AEI-45		2K×8	LCC	-40° to +85°C	Standard	450ns	
X2816AEM-35		2K×8	LCC	-55° to +125°C	Standard	350ns	
X2816AEM-45		2K×8	LCC	-55° to +125°C	Standard	450ns	
X2816AEM-45		2K×8	LCC	-55° to +125°C	883B	350ns	
X2816AEM-45		2K×8	LCC	-55° to +125°C	883B	450ns	
X2864AD		8K×8	Cerdip	0° to +70°C	Standard	300ns	
X2864AD-35		8K×8	Cerdip	0° to +70°C	Standard	350ns	
X2864AD-45		8K×8	Cerdip	0° to +70°C	Standard	450ns	

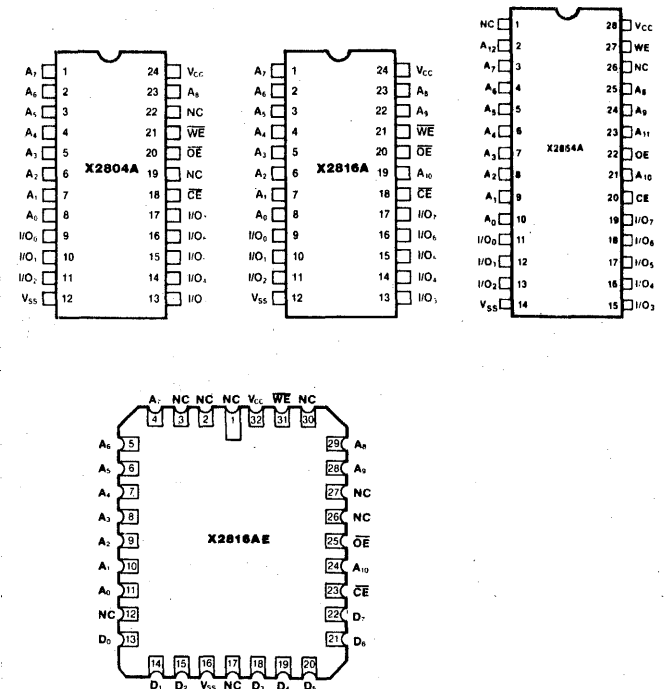
\* Serial Interface

## PIN CONFIGURATIONS

### NOVRAM



### E<sup>2</sup>PROM







**X2816A 2K x 8 Bit Electrically Erasable PROM**  
**X2804A 512 x 8 Bit Electrically Erasable PROM**

**5 Volt Programmable E<sup>2</sup>PROMs**

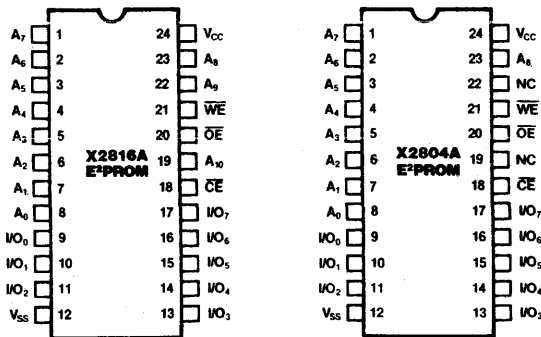
- Simple Byte Write Operation
  - No High Voltages Necessary
  - Single TTL level  $\overline{WE}$  Signal Modifies Data
  - Internally Latched Addresses and Data
  - Automatic Write Time-out
  - Noise Protected  $\overline{WE}$  Pin
- Conforms to JEDEC Byte-wide Standard
- Reliable N-Channel Floating Gate MOS Technology
- Single 5-Volt Supply
- Byte Write Time: 10ms Max.
- Fast Access Time: 300ns Max.
- Low Power Dissipation
  - Active Current X2816A: 110mA Max.  
X2804A: 80mA Max.
  - Standby Current: 50mA Max.

The Xicor X2816A (16,384 bits) and X2804A (4,096 bits) are electrically erasable programmable read-only memories (E<sup>2</sup>PROMs) with unprecedented ease-of-use features. Xicor E<sup>2</sup>PROM data can be modified using simple TTL level signals and a single 5-volt power supply. In addition, Xicor E<sup>2</sup>PROMs are operationally and pin compatible with existing 2K x 8 byte-programmable E<sup>2</sup>PROMs which require an additional high voltage power supply for programming. (See *optional* high voltage programming compatible mode.) Writing data in Xicor E<sup>2</sup>PROMs is analogous to writing data in a static RAM. A 200ns TTL low level signal to the  $\overline{WE}$  pin initiates a byte write operation which is automatically timed out in a maximum of 10ms. Since addresses and data are internally latched, Xicor E<sup>2</sup>PROMs free the system for other tasks during the 10ms period, such as programming other Xicor E<sup>2</sup>PROMs. In addition to byte modification capability, a 10ms total chip erase feature is provided.

Xicor E<sup>2</sup>PROMs use a 2-line control architecture,  $\overline{CE}$  and  $\overline{OE}$ , to eliminate bus contention in a system environment. A power down mode is featured. In the standby mode, power consumption is reduced by 55% without increasing access time. The standby mode is achieved by applying a  $\overline{CE}$  high signal.

The X2816A and X2804A are fabricated with the same reliable n-channel floating gate MOS technology used in Xicor's popular 5-Volt programmable NOVRAM memories.

**PIN CONFIGURATIONS 24 PIN DIP .600"**



**PIN NAMES**

A <sub>0</sub> -A <sub>10</sub>	ADDRESS INPUTS
I/O <sub>0</sub> -I/O <sub>7</sub>	DATA INPUTS/OUTPUTS
$\overline{CE}$	CHIP ENABLE
$\overline{OE}$	OUTPUT ENABLE
$\overline{WE}$	WRITE ENABLE
V <sub>CC</sub>	+5V
V <sub>SS</sub>	GROUND
NC	NO CONNECT

**MODE SELECTION**

Standard Xicor 5V-Programmable Mode

$\overline{CE}$	$\overline{OE}$	$\overline{WE}$	MODE	I/O	POWER
H	X	X	Standby	High Z	Standby
L	L	H	Read	D <sub>OUT</sub>	Active
L	H	L	Byte Write	D <sub>IN</sub>	Active
L	H	H	Read and Write Inhibit	High Z	Active

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 Preliminary Information Sheet  
 May 1983, Stock No. 200-012

MEMORY

64K

X2864A

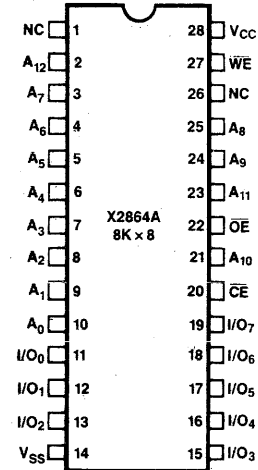
8K x 8 Bit  
Electrically Erasable PROM

5 Volt E<sup>2</sup>PROM

### FEATURES

- Simple Byte and Page Write
  - Single TTL Level  $\overline{WE}$  Signal
  - Latched Address and Data
  - Automatic Internal Erase
  - Automatic Write Timing
  - Automatic Page Write
  - DATA Polling Verification
  - Optional Chip Erase
- Enhanced Write Protection
- Single 5 Volt Supply
- Byte or Page Write: 5 ms Typical
  - Effective 300  $\mu$ sec/Byte Write
  - Chip Rewrite—2.6 sec
- Fast Access Time: 350 ns
- Power: 110 mA—Active Current  
50 mA—Standby Current
- JEDEC Approved Byte-Wide Pinout

### PIN CONFIGURATION



### DESCRIPTION

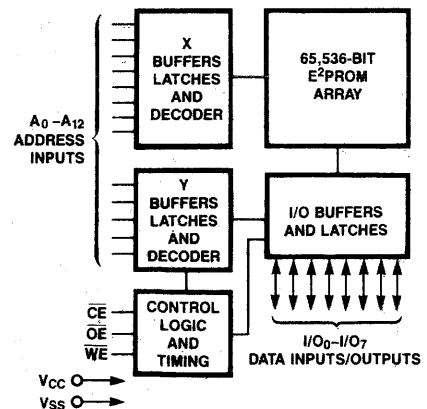
The Xicor X2864A is an electrically erasable programmable read-only-memory (E<sup>2</sup>PROM) with unprecedented ease-of-use features. Xicor E<sup>2</sup>PROMs can be modified using simple TTL level signals and a single 5 volt power supply. In addition, the X2864A is operationally and pin compatible with existing 8K x 8 EPROMs which require high voltage programming and ultraviolet exposure erasing. Writing data in Xicor E<sup>2</sup>PROMs is analogous to writing data in a static RAM. A 200 ns TTL low level signal to the  $\overline{WE}$  pin initiates a byte write operation which is automatically timed out in 5 ms. Since addresses and data are internally latched, Xicor E<sup>2</sup>PROMs free the system for other tasks during the 5 ms period, such as programming other Xicor E<sup>2</sup>PROMs.

A 16-Byte Page Write allows data to be written at an effective rate of 300  $\mu$ sec/byte or 2.6 seconds to rewrite the entire chip. An optional chip erase feature is also included.

The X2864A also features  $\overline{DATA}$  Polling, which enables the E<sup>2</sup>PROM to signal the processor that a write operation is complete without requiring the use of any external hardware.

The X2864A is fabricated with the same reliable N-Channel floating gate MOS technology used in the popular 5 volt programmable Xicor NOVRAMs.

### FUNCTIONAL DIAGRAM



### MODE SELECTION

$\overline{CE}$	$\overline{OE}$	$\overline{WE}$	Mode	I/O	Power
L	L	H	Read	D <sub>OUT</sub>	Active
L	H	L	Write	D <sub>IN</sub>	Active
H	X	X	Standby and Write Inhibit	High Z	Standby
X	L	X	Write Inhibit	—	—
X	X	H	Write Inhibit	—	—
L	V <sub>OE</sub>	L	Chip Erase	D <sub>IN</sub> = H	Active



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Circuits Magazine**

# INTRODUCTION TO PROM PROGRAMMERS

This section describes instruments used to enter assembly language code into programmable devices such as PROMs, EPROMs, EEPROMs, EAROMs, PALs, PLAs and MPUs. Equipment listed includes programmers designed for system development, production and field-service applications. Manufacturers are sequenced alphabetically, and each system is characterized by pertinent selection factors. These parameters include programmed device types, type of display, personality modules, edit functions, programming checks, self-test features and types of interface ports.

Some manufacturers assign different model numbers to units having capabilities beyond those listed for the model designated at the top of the Master Selection Guide. In these cases the units are covered under "Comments".

## Detailed Product Information provided by:

Oliver Advanced Engineering	4401
Structured Design	4402
Sunrise Electronics	4403

The manufacturers listed above have provided detailed information on their latest and most significant products.

# IC MASTER

## PROM PROGRAMMERS

Application	Capabilities	Programmed Devices	Display	Maximum PROM Array (bits)	Personality Modules	Maximum Gang Capacity (Sockets)	Edit Functions
Development	Program, copy, compare, read	EPROM	CRT, LED	16Kx8	Not Required		Insert, delete
Production	Copy	EPROM, EEPROM	LED	32Kx8	Software based	16	
Development, production, field service	Program, copy, compare, read, erase	EPROM	LED	32Kx8	Not required, software-based	1	Insert, move, delete, nibble swap, memory map
Production	Program, copy, compare	Bipolar PROM	Lamps	4Kx8	Dedicated	1	
Development, field service, Production	Program, copy, compare, read, erase	NMOS, CMOS, MPU, EEPROM, EPROM	7-Segment LED(4)	23Kx8 expandable	Software programmable	1	Via Host System
Development, field service	Program, read	Bipolar PROM	Lamps, digit switch	4Kx8	Dedicated		
Development, production, field service	Program, copy, compare, read, test	FPLA: Signetics 82S100, 82S101, 82S106, 82S107	LED		Dedicated		Insert, delete
Development, field service, production	Program, copy, compare, read	Bipolar PROM (8 x 32)	LED	32x8			Insert, delete
Development	Program, copy, compare, read, test	Seeq 72720 (T.I. 7000 with EEPROM)	User terminal	up to 8Kx8		1	Line editing or any word processor on host computer
Development	Program, copy, compare, read, test, erase	EPROM, EEPROM	CRT	32Kx8	Dedicated		
Development, Incoming Test	Program, copy, compare, read, test	8748, 8749	User terminal	1Kx8 8748, 2Kx8 8749		1	Line editing or any word processor on host computer
Development, incoming test	Program, copy, compare, read, test	8751, 8752	User terminal	2Kx8 8751, 4Kx8 8752		1	Line editing or any word processor on host computer
Development	Program, copy, compare, read	EPROM, EEPROM, MPU	LED	8Kx8	Dedicated (Universal NMOS and Intel single-chip microcomputers)	1	Insert, delete, write, fill
Development, field service	Program, copy (load), compare (verify), read, test, erase	Bipolar (fuse-link and vertical-fuse) PROM, CMOS, PROM, EPROM, EEPROM, EAROM, MPU	16-character fluorescent	256Kx8	24-pin skinny-wide and 40-pin MPU socket adapters; no modules required	1	Insert, move, delete, nibble swap, memory map, split, shuffle, clear, complement, fill RAM
Development, field service	Program, copy (load), compare (verify), read, test, erase	CMOS, PROM, EPROM, EEPROM, EAROM, MPU	16-character fluorescent	256Kx8		1	Insert, move, delete, nibble swap, memory map, split, shuffle, clear, complement, fill RAM

PROM PROGRAMMERS

Programming Checks	Self-Test	Prompting	Buffer Memory	I/O Interface	Comments	Model	Source	Line
Blank, verify		Audible		RS232 Serial, selectable baud rate	Operates with Z80/8080 CP/M. Utility software supplied to partition long object or HEX files into required EPROM length and split 16-bit files into upper/lower 8-bit files.	UP8	Advent	
Blank, checksum, illegal bit, verify	Keyboard, display, switches	Visual/audible				CopyROM3716	Citel	
Blank, checksum, illegal bit, verify	Keyboard, display, switches, buffer, memory	Visual/audible	8Kx8 expandable	Parallel TTL, RS232 serial		System37	Citel	
Illegal bit, Vcc sensitivity		Visual				PD2000S	Curtis	
Blank, checksum, illegal bit, verify, on current	Status of programming levels	Visual/audible	64	RS232 serial selectable Baud rates 110-9600	Programs 4K 3-supply through 25gK 1-5 supply and beyond. Stand-alone duplicator or complete remote control by host computer.	PD5000E	Curtis	5
Vcc sensitivity		Visual				PM3000S	Curtis	
Verify		Visual				PR-100A	Curtis	
Blank, checksum, illegal bit, verify	Diagnostic & calibration	Visual/Audible	32 bytes	RS5332 110-9600 Baud	Handles Signetics, TI, NSC, Harris, AMD, MMI, Intersil.	PR5200B	Curtis	
Checksum verify bytes		Visual	User system	300, 1200, 2400 baud RS 232 link to host computer	IBM PC cross assemblers with downloader available. Programmer board connects to any RS-232 port and utilizes user computer terminal and power supply.	CYP-7000	Cybernetic	
Blank, checksum, verify		Visual/Audible	16Kx8	RS232 Serial, Selectable Baud Rate		CYP275	Cybernetic	10
Checksum verify bytes		Visual	User system	300, 1200, 2400 baud RS 232 link to host computer	IBM PC cross assemblers with downloader available. Programmer board connects to any RS-232 port and utilizes user computer terminal and power supply.	CYP8048/9	Cybernetic	
Checksum verify bytes		Visual	User system	300, 1200, 2400 baud RS 232 link to host computer	IBM PC cross assemblers with downloader available. Programmer board connects to any RS-232 port and utilizes user computer terminal and power supply.	CYP8051	Cybernetic	
Blank, checksum, illegal bit, verify	Keyboard, display, switches, buffer memory, automatic power-up, program ROM, status of programming electronics	Visual	2Kx8	RS232 serial, selectable baud rate (110-9600 baud)		20B Universal NMOS Memory Programmer	Data I/O	
Blank, checksum, illegal bit, verify, overcurrent check, backward device, 2-pass verify with variable	Keyboard, display, switches, buffer memory, automatic power-up, program ROM, status readout of programming electronics	Visual	32Kx8	RS232 serial, current loop, selectable baud rate (50-19,200)	Portable programming system; includes UV erasing lamp and removable lid for storing accessory.	22A Personal Programmer	Data I/O	
Blank, checksum, illegal bit, verify, overcurrent check, backward device, 2-pass verify with variable Vcc	Keyboard, display, switches, buffer memory, automatic power-up, program ROM, status of programming electronics	Visual	32Kx8	RS232 serial, current loop, selectable baud rate (50-19,200)	Portable MOS-only programming system; includes UV erasing lamp and removable lid for accessory storage.	22B Personal Programmer	Data I/O	15

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## PROM PROGRAMMERS

Application	Capabilities	Programmed Devices	Display	Maximum PROM Array (bits)	Personality Modules	Maximum Gang Capacity (Sockets)	Edit Functions
Development, field service	Program, copy (load), compare (verify), read, test	Bipolar PROM, CMOS, PROM, EPROM, EEPROM, EAROM, PAL, PLA, MPU, FPGA, FPLA, FPRP, FPLS, prog. I/O port, PMUX	16-character fluorescent	256Kx8 (UniPak2)	Software selectable Paks (UniPAK 2, MOSPak, LogicPak, Gang Pak, Handler UniPak)	8	Insert, move (block), delete, nibble swap, memory map, split, shuffle, clear, complement, fill RAM
Production	Program, copy (load), compare (verify), read, test, EEPROM cycling test, electronic device identification	MOS and CMOS PROM, EPROM, EEPROM, intelligent algorithms for EPROMs	16 character fluorescent	16Kx8, expandable	Software selectable	20	Complement RAM, fill RAM, 16-bit mode
Development, production	Program, Set program, copy (load), compare (verify), read, test, EEPROM cycling test, electronic device identification	MOS and CMOS, EPROM, EEPROM, intelligent algorithms for EPROMs	16 character fluorescent	16Kx8	Software selectable	20	Insert, delete, I/O offset, begin RAM, fill RAM, complement RAM, 16-bit mode
Development, production, field service	Program, copy, compare, read, test, stimulate	CMOS, PROM, EPROM, EEPROM	LED Hexadecimal	32Kx8	Dedicated, generic		Insert, move, search, split/shuffle, memory map
Production, field service	Program, copy, compare, read, test	CMOS, PROM, EPROM	LED	4Kx8	Gang, generic gang	16	
Production, field service	Program, copy, compare, read, test, erase	CMOS, PROM, EPROM, EEPROM	LED	16Kx8	Gang, generic gang	10	
Development, production, field service	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS, PROM, EPROM, EEPROM, EAROM, IFL, PAL, PLA, MPU	CRT	63Kx8	Generic		Insert, move, delete, nibble swap, memory map, complement, byte, split, shuffle, search
Development, production	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS, PROM, EPROM, EAROM, PAL, PMUX, MPU	LED, hexadecimal	50Kx8 (add field 64Kx8BMS)	Dedicated, generic		Move, nibble swap, memory map, complement
Development, production, field service	Program, copy, compare, read	Bipolar PROM, EPROM, MPU	LED	16Kx8	Dedicated, generic, gang	8	Insert, delete, nibble swap
Development, field service	Program, copy, compare, read, test, erase	CMOS, PROM, EPROM, EEPROM, EAROM	LED	32Kx8	Dedicated		Insert, move, delete, memory map
Development, field service	Program, copy, compare, read, test, erase, modify	Bipolar PROM, CMOS, PROM EEPROM			Not required		Move, compare, replace
Development, field service	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS PROM, EPROM, EEPROM, EAROM, PAL	LED	32Kx8	Generic & software programmable		Insert, move, delete, nibble swap, memory map, search

PROM PROGRAMMERS

Programming Checks	Self-Test	Prompting	Buffer Memory	I/O Interface	Comments	Model	Source	Line
Blank, checksum, illegal bit, verify, overcurrent, backward device, 2-pass verify with variable Vcc	Keyboard, display, switches, buffer memory, automatic power-up, program ROM, status of programming electronics	Visual/audible	16Kx8 (standard), 64Kx8, (optional)	RS232 serial, current loop, selectable baud rate (50-19,200 baud)	Edit in hexadecimal, octal or binary. English-language prompts, menus and error messages. Remote control and 26 development-system format translators are standard.	29A Universal Programmer	Data I/O	1
Blank, checksum, illegal bit, verify, overcurrent, backward device, mislocated device	Keyboard, display, switches, buffer memory, automatic power-up, program ROM, status of programming electronics	Visual/audible	32Kx8	RS232 serial, selectable baud rate (110-9600 baud)	Remote control and 26 data translation formats are standard. M120A controls 200A Data Control Unit for program storage.	M120A Gang Programmer	Data I/O	
Blank, checksum, illegal bit, verify, overcurrent, backward device, mislocated device, variable Vcc on verification	Keyboard, display, switches, buffer memory, automatic power-up, program ROM, status of programming electronics	Visual/audible	128Kx8, 256Kx8	RS232 serial (2), selectable baud rate (110-9600 baud)	Can segment programs and load each segment into different PROMs simultaneously. Remote control and 26 data translation formats are standard. M121A controls 200A Data Control Unit for program.	M121A Gang Programmer	Data I/O	
Blank, illegal bit, checksum, verify	Automatic with power-up	Visual/audible	8Kx8 expanded to 32Kx8 data retention with power off	RS232 serial port, selectable baud rate to 19.2K	Interfacing formats: Intel Hex; Motorola exorciser; ACSII-Hex Space; ACSII-BNPF & manual format. EPROM/EEPROM simulation capability	EP-804	Digelec	
Blank, verify		Visual/audible	4Kx8	RS232 serial, current loop, selectable baud rate	Can program two different patterns simultaneously.	GP-16 Gang Programmer	Digelec	5
Blank, checksum, illegal bit, verify, parametric test	Device Selection vs Pin Out	Visual/audible				GP-810 Gang Programmer	Digelec	
Blank, checksum, illegal bit, verify, continuity, parametric check	Buffer memory, program voltages	Visual/audible	64Kx8, expandable to 512Kx8 battery backup	Parallel TTL, RS232 serial, current loop, selectable baud rate	No calibration required. Software/Menu driven machine.	UP-803	Digelec	
Blank, verify, continuity, window check		Visual/audible	4Kx8, 8Kx8, expandable, battery backup	Parallel TTL, RS232 serial, current loop, selectable baud rate	12 interfacing formats; 4 data entry keyboard formats	UPP-801	Digelec	
Blank, checksum, illegal bit, verify	Display, buffer memory	Visual/audible	16Kx8	RS232 serial, current loop, selectable baud rate, 110 to 9600 baud	Unit is equipped with three fast programmable power supplies (100ns/V) for programming voltages	RP400-S	Elind	
Blank, checksum, illegal bit, verify	Buffer memory	Visual/audible	64Kx8	RS232 serial, + TTL serial	The RP600 is a dedicated EPROM Programmer, the selection is made by software. The unit is particularly suited for field service. The unit can program all EPROMs available at the moment.	RP600	Elind	10
Verify						H3000	Hughes	
Blank, checksum, illegal bit, verify	Keyboard, display, buffer memory module	Visual	32Kx8	RS232 serial, current loop, selectable baud rate	PROM Emulation Module option allows user to test and debug programs before committing to PROMs.	IM1010 Universal Programmer	IMS	

Bold face indicates additional data is provided on the page noted.



PROM PROGRAMMERS

Application	Capabilities	Programmed Devices	Display	Maximum PROM Array (bits)	Personality Modules	Maximum Gang Capacity (Sockets)	Edit Functions
Production, development	Program, copy, compare, read, test, erase	EPROM, EEPROM, CMOS, EPROM	LED	16Kx8, 32Kx8	Software programmable	16	Move, delete, memory map, I/O offset, fill RAM
Development, production	Program, copy, compare, read, erase (E2 only)	Bipolar PROM, EPROM, EEPROM, MPU	LED	32Kx8bits	Generic, gang, generic gang	8	Move*, nibble swap* n byte swap*, memory map*, enter
Development, production	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS PROM, EPROM, EEPROM, EAROM, PAL, PLA, MPU, EPLA, FDRP, EPGA, FPLS, PMUX	Fluorescent	32Kx8	Dedicated, generic, gang, generic gang	8	Set, insert, move, delete, nibble swap, memory map, fill, invert, 16-bit split/shuffle, logical AND/OR/XOR, string search
Development, production, field service	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS PROM, EPROM, EEPROM, EAROM, PAL, PLA, FPLA, FPRP, FPGA, FPLS, PMUX	LED	32Kx8	Dedicated, generic, gang, generic gang	8	Insert, move, delete, nibble swap, memory map, fill, invert
Development, production, field service	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS PROM, EPROM, EEPROM, EAROM, PAL, PLA, MPU, FPLA, FPLS, PMUX	LED	32Kx8	Dedicated, generic, gang, generic gang	8	Set, insert, move, delete, nibble swap, memory map, fill, invert
Development, production, field service	Program, copy, verify, read, test, erase, help, odd/even	Bipolar PROM, CMOS PROM, EPROM, EEPROM, EAROM, PLA	CRT via terminal	8Kx8	Dedicated	1	Insert, move, delete, save, memory map, change
Development, production, field service	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS PROM, EPROM, EEPROM, EAROM, PAL, PLA, MPU, All programmable devices	CRT Interfaces with Remote printer		None—(completely software based)!!	1	Insert, move, delete, nibble swap, memory map, on screen cursor-controlled editing
Production	Program, copy, compare, read, test	EPROM	LED	32Kx8	Generic gang	16 (32 with satellite)	None allowed
Production	Program, copy, compare, read, test, erase, dc parameters	EPROM, EEPROM, EAROM	LED, ASCII	256Kx8	Generic gang	18 (36 with satellite)	None allowed
Production	Program, copy, compare, read, test, erase, dc parameters	EPROM, EEPROM, EAROM	LED, ASCII	256Kx8	Generic gang	20 (40 with satellite)	None allowed
Development, production	Program, copy, compare, read	EPROM	LED	2764	Dedicated	1	Insert, move
Development, production, field service	Copy, compare	Bipolar PROM, CMOS PROM, EPROM, EEPROM, PLA	LED	64Kx8	Dedicated, generic, gang, generic gang	8	Insert, move, delete, nibble swap, memory map
Development, production, field service	Program, copy, compare, read	Bipolar PROM, CMOS PROM, EPROM, EEPROM	LED	64Kx16	Dedicated, generic, gang, generic gang	8	Insert, move, delete, nibble swap, memory map

PROM PROGRAMMERS

Master Selection Guide

PROM PROGRAMMERS

Programming Checks	Self-Test	Prompting	Buffer Memory	I/O Interface	Comments	Model	Source	Line
Blank, checksum, illegal bit, verify, misaligned or upside-down device, broken or bent pin, overcurrent	Keyboard, display, switches, buffer memory, automatic power reset, calibration, programming electronics	Visual/audible	64Kx9 standard; 128Kx9, 256Kx9 and 512Kx9	RS232 serial, selectable baud rate	Simultaneous set programming ability, electrically isolated programming sockets, Computer control software automates EPROM programming, intelligent Fast algorithms supported.	IM3016 Multi-Master Programmer	IMS	1
Blank, checksum, illegal bit, verify, reverse socket detection	Buffer memory, motherboard, power supply, module	Visual/audible*	16Kx8, expandable to 32K; virtual buffer*	RS232 serial, automatic baud rate selection	*iUPS software package feature. Host software driver also provides file manipulation, file mapping, held file, screen editing, data manipulation and other features.	UP-200/201	Intel	
Blank, checksum, illegal bit, verify at high and low Vcc limits	Keyboard, display, switches, buffer, firmware integrity	Visual/audible	8Kx8, expandable to 32Kx8	Parallel TTL, RS232 serial, current loop, selectable baud rate (50 to 38400 baud), IEEE-488	Build-in uv EPROM eraser with keyboard programmable timer. Remote controllable via RA232 and IEEE-488 ports. Error-tolerant communications protocol for long distance telephone transfers. Compatible with MPP-80S modules.	EPP-80	Kontron	
Blank, checksum, illegal bit, verify at high and low Vcc limits	Keyboard, display switches, buffer memory	Visual/audible	4Kx8, 8Kx8, 16Kx8, 32Kx8 plus mass storage	RS232 serial, selectable baud rate	Build-in uv source, IC handler interface. MPP 80SAM has acoustic coupler and modem.	MPP-80	Kontron	
Blank, checksum, illegal bit, verify at high and low Vcc limits	Keyboard, display switches, buffer memory, firmware integrity	Visual/audible	8Kx8, expandable to 32Kx8	RS232 serial, current loop, selectable baud rate (50 to 38400 baud)	Build-in uv EPROM eraser. IC handler interface. Remote controllable via RS232 port from terminal or host computer. MPP-80SAM has build-in acoustic coupler, modem and error-tolerant communications protocol.	MPP80S	Kontron	5
Blank, illegal bit, verify		Visual/audible	8Kx8, expandable	Proprietary		MG8PPS	Motorola	
Blank, checksum, illegal bit, verify, matrix test	Keyboard, buffer memory, leakage test on socket	Visual/audible menu-driven	64Kx8 Std, expandable to 256Kx8	Dual RS-232	Five versions available, depending on programming capabilities needed. Fully expanded, Datatap uses 40 separate high-voltage drivers with a library of over 1000 programmable devices, programmed from one universal socket!!	Datatap*	Oliver	
Blank, illegal bit, verify, matrix test*	Keyboard, switches, leakage test on socket matrix	Visual/audible			*Matrix test checks both data and address lines for ESD (electrostatic discharge) damage. Data lines are checked for min. sink and source current capability during both blank and verify tests. 100-240 Vac operation.	UPP-2700	Oliver	
Blank, checksum, illegal bit, verify, Vcc level tests, matrix test	Keyboard, switches, leakage test on socket matrix	Visual/audible, including voice option	2 ports for label printer and voice enunciator	UL-Listed, short circuit protection, power-fail restart capability		UPP-28000-ZIF	Oliver (4401)	
Blank, checksum, illegal bit, verify, Vcc level tests, matrix test	Keyboard, switches, leakage test on socket matrix	Visual/audible, including voice option		2 ports for label printer and voice enunciator		UPP-28000 Scope	Oliver (4401)	10
Blank, illegal bit, verify		Visual	None	RS232 serial, selectable baud rate		Gangbuster	Onset	
Blank, verify, at high and low Vcc limits, checksum	Keyboard, display, switches, buffer memory	Visual/audible	8Kx8, 16Kx8, expandable, battery backup CMOS	Parallel TTL, RS232 serial, selectable baud rate	Two year warranty. Separate master and copy sockets on personality modules (can duplicate directly from master to copy socket). Fixed Vpp power supplies do not require adjustment.	M910AControl Unit	Pro-Log	
Blank, checksum, illegal bit, verify at high and low Vcc limits	Keyboard, display, switches, buffer memory	Visual/audible	8Kx8, 16Kx8, expandable, battery backup CMOS	Parallel TTL, RS232 serial, selectable baud rate	Two year warranty. Separate master and copy sockets on personality modules (can duplicate directly from master to copy socket). Fixed Vpp power supplies do not require adjustment.	M980Control Unit	Pro-Log	

Bold face indicates additional data is provided on the page noted.

# IC MASTER

## PROM PROGRAMMERS

Application	Capabilities	Programmed Devices	Display	Maximum PROM Array (bits)	Personality Modules	Maximum Gang Capacity (Sockets)	Edit Functions
Development, production, field service	Program, copy, compare, read, test, erase	Bipolar PROM, CMOS PROM, EPROM, EEPROM, PAL, MPU	LED	64Kx16	Dedicated, generic, gang, generic gang	8	Insert, move, delete, nibble swap, memory map, split/interweave (4 and 8 bit), invert
Development, production, field service	Program, copy, compare, read, test	PAL	KED	2K	Dedicated	1	Insert, move, delete, memory map
Development, production	Program, copy, compare, read, test	PAL					Insert, delete, memory map
Production	Program, copy, compare, read, test for device programmability, erase EEPROM, device software	Bipolar PROM, CMOS PROM, EPROM, EEPROM, PAL, FLPA, MPU	14 character alphanumeric LED	256K bits	Slave for FPLA, PAL AIM devices and gang EPROM programming	24 using 2 gang slaves	Peek, poke, move, insert, clear DATA RAM to state, odd/even byte separate/swap, list, find, checksum and selected range programming
Production	Gang program, copy, compare, read, test for device programmability, EPROMS, EEPROMS and MPU's and erase EEPROMS. Devices software selected.	CMOS PROM, EPROM, EEPROM, MPU, PAL, FPLA	14 character alphanumeric LED	256K bits	Slave for FPLA, PAL AIM devices and gang EPROM programming	36 using 2 gang slaves	Peek, poke, move, insert, clear DATA RAM to state, odd/even byte separate/swap, list, find, checksum and selected range programming
Production	Gang program, copy, compare, read, test for device programmability.	8748, 8748H, 8749H, 8741, 8755A, 8751	LED	256K bits	Slave for FPLA, PAL AIM devices and gang EPROM programming	24 using 2 gang slaves	Peek, poke, move, insert, clear DATA RAM to state, odd/even byte separate/swap, list, find, checksum and selected range programming
Development, field service	Program, copy, compare, read, test, erase EEPROM and simulate. Devices software selected	EPROM, EEPROM, Intel and Motorola MPU's	8-character LED	128K bits			Peek, poke, move, clear DATA RAM to state, odd/even byte separate/swap list, checksum and selected range programming
Development, production	Read, program, load from disk, verify, verify Blk, store to disk, program chip set	EPROM, bipolar PROM, CMOS, MPU, EEPROM, EAROM, PAL, PLA, FPGA, FPLA, FPLS, PMUX, FPRP, prog I/O port	80 characters x 24 line CRT	64Kx8		9 (40 pin)	Change, move, fill, complement, word order/-size, memory map
Development, field engineering	Read, program, load from disk, verify, verify Blk, store to disk, erase	EPROM, bipolar PROM, CMOS, MPU, EEPROM, EAROM, PAL, PLA, FPGA, FPLA, FPLS, PMUX, FPRP, prog I/O port	80 characters x 24 line CRT	64Kx8		1	Change, move, fill, complement, word order/-size, memory map

## PROM PROGRAMMERS

Programming Checks	Self-Test	Prompting	Buffer Memory	I/O Interface	Comments	Model	Source	Line
Blank, checksum, illegal bit, verify, backward PROM, shorted address line, overload (Vcc and Vpp)	Keyboard, display, switches, buffer memory, programmable algorithms	Visual/audible	8Kx8, 16Kx8, expandable, battery backup, CMOS	Parallel TTL, RS232 serial, selectable baud rate, TTY		System90	Pro-Log	1
Blank, verify, function test, continuity	Display, switches, buffer memory	Visual	2Kx8	RS232 serial, selectable baud rate	PALs are designed simulated programmed and functionally tested on the PAL BURNER. PAL design specifications may be down loaded from another computer or stored in the on-board storage media. Programs all MMI, AMD, TI, and NATIONAL 20 and 24 pin PALs	SD1000	StrucDes	(4402)
Illegal bit, verify			2Kx8	RS232 Serial, selectable baud rate		SIMPALINK	StrucDes	
Blank, programmability checksum, illegal bit, verify	Keyboard, display, switches, DATA RAM, automatic voltage and timing recalibration before and during programming	English word, audible	16Kx8 expandable to 64Kx8	Dual independent RS232 serial ports selectable formats parity, baud rates 110-19.2K, terminal and computer control	Z-2000 has same features as Z-1000 except memory is 64Kx8 and has interfaces for 8 inch dual disc drives and IEEE488.	Z1000/Z2000	Sunrise	(4403)
Blank, programmability checksum, illegal bit, verify	Keyboard, display, switches, DATA RAM, automatic voltage and timing recalibration before and during programming	English word, audible	16Kx8 expandable to 64Kx8	Dual independent RS232 serial ports selectable formats parity, baud rates 110-19.2K, terminal and computer control	Z-2400 has same features as Z-1200 except memory is 64Kx9 and has 8 inch dual disc drives and IEEE-488 included.	Z1200&Z2400 Gang Programmers	Sunrise	(4403)
Blank, programmability checksum, illegal bit, verify	Keyboard, display, switches, DATA RAM, automatic voltage and timing recalibration before and during programming	English word, audible	16Kx8 expandable to 64Kx8	Dual independent RS232 serial ports selectable formats parity, baud rates 110-19.2K, terminal and computer control, 8-bit parallel port	Z-2448 has same features as Z-1248 except memory is 64Kx8 and it has 8-inch dual disc drives and IEEE-488 included.	Z1248	Sunrise	(4403)
Blank, checksum, illegal bit, verify	Keyboard, DATA RAM, simulator, programming sockets, display	Visual/ audio	8Kx8 expandable to 16Kx8 battery operated	RS232 serial, selectable formats, parity, baud rates 110-19.2K, terminal and computer control with selectable communications	PROM data display simultaneous with RAM data. ZAP48 has same features as plus programming of Intel's 8748, 8749, 850, 8741, 8755A and 8751MPU's. ZAP68 has same features as ZAP80 plus programming of Motorola's 68701 and 68705 MPU's.	ZAP80/ZAP48/ZAP68	Sunrise	(4403)
Verify blank, verify programmable block check, verify overcurrent, backward, wrong chip type, threshold voltage margins, mislocated device	Registers, memory, drivers, power usage, auto-calibration to .1% on current and voltage	Visual LED's, visual CRT	Virtual paged to disk—no limit	RS232 serial, selectable baud rate (110-9600) Parallel printer port	New chips are supported through software distribution on diskette. Statistics gathering gives printout of programming activity and yield. Automatic labeling support standard. Single button repeat programming.	PG1140-X	Varix	
Verify blank, verify programmable block check, verify overcurrent, backward, wrong chip type, threshold voltage margins, mislocated device	Registers, memory, drivers, power usage, auto-calibration to .1% on current and voltage	Visual LED's, visual CRT	Virtual paged to disk—no limit	RS232 serial, selectable baud rate (110-9600) Parallel printer port	New chips are supported through software distribution on diskette. Customer may write software to program or test his own proprietary chips.	SP0300-X	Varix	

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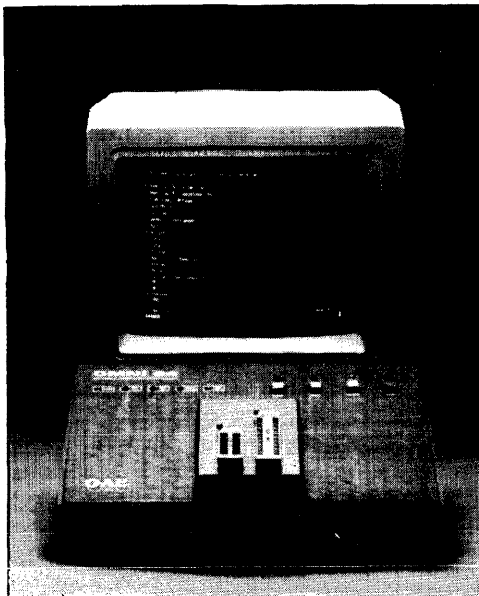


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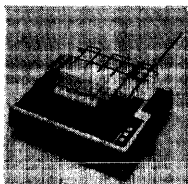
- single chip microprocessors, PALs, FPLAs, FPLSs, FPGAs, and even diode arrays.
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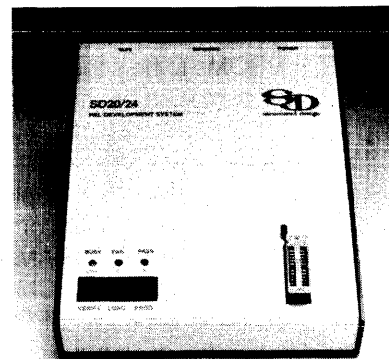
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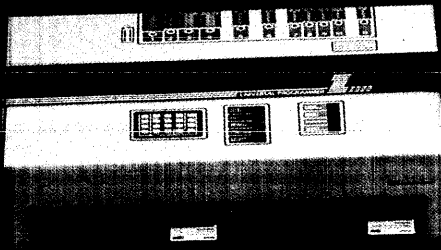
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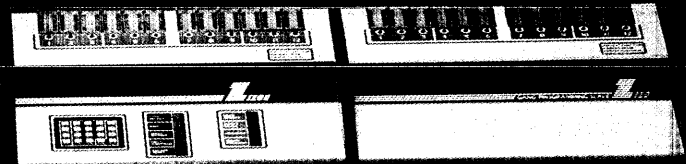
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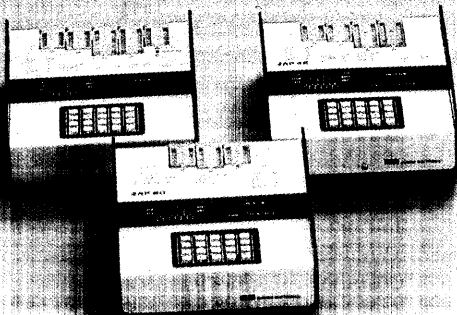
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## Z-120 Gang Programmer

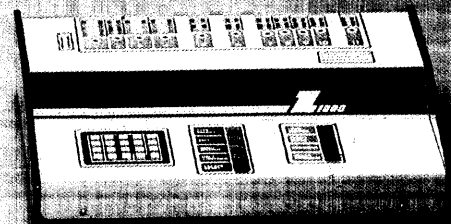
Z-120 gang programmer for EPROMs and microcomputers. You can add Z-120 gang programmer to multiply your throughput and program Intel's 87XX family of microcomputers. Z-1248 programmer or Z-148 slave.



## Zap Series EPROM/EEPROM/MPU Programmers

The ZAP SERIES of portable, battery operated programmers are designed for your development and field service needs.

These handy units program EPROMs and EEPROMs thru 27128, Intel and Motorola MPUs, use fast algorithms and have built-in PROM simulators. They also feature keystroke device selection, 16K x 8 DATA RAM, a sophisticated editor, 110 to 19.2K baud RS-232 ports and terminal control.



## Z-1000/1000B Universal Programmers

The Z-1000 and 1000B program your EPROMs, bipolar PROMs, PALs, FPLAs and Intel microcomputers. Prompted device selection, fast algorithms and easily upgraded software let you program almost any device available now or in the future. These units are slave expandable for EPROM and MPU gang programming.

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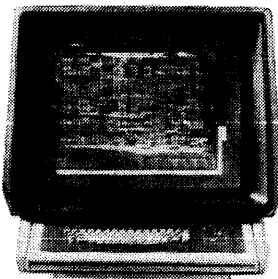
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# INTRODUCTION TO CUSTOM/SEMICUSTOM CIRCUITS

This section describes services and products of companies supplying digital, linear and combined digital/linear custom circuits. Listed alphabetically in the Master Selection Guide, the companies are characterized by descriptions of design services, production facilities, process technologies and testing capability. Options for user contribution to custom-circuit projects are outlined by listing the vendor's preferred level of user input which can range anywhere from system concept to a known good device.

Semicustom Gate Arrays, Linear and Linear/Digital Arrays, Standard Cells, and Programmable Logic are covered in detail appropriate for locating devices or suppliers. In the semicustom area, performance parameters are circuit dependent and definitions are not standardized; therefore some are more useful in comparisons within a manufacturer's line than in comparisons between manufacturers. Frequently the manufacturer's data provides additional capability information that will assist in source selection.

## Detailed Product Information provided by:

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The manufacturers listed above have provided detailed information on their latest and most significant products.

# EINFÜHRUNG KUNDEN- SPEZIFISCHE ICs

Dieser neue Abschnitt beschreibt Dienstleistungen und Produkte von Herstellern, die digitale, lineare und kombinierte digital/lineare kundenspezifische Schaltkreise anbieten. Die Firmen werden in alphabetischer Reihenfolge genannt und anhand ihrer Fähigkeit zum Schaltkreis-Entwurf, ihrer Produktionsanlagen, Prozeßtechniken und Testmöglichkeiten beschrieben. Die verschiedenen Möglichkeiten des Anwenders im Hinblick auf dessen Angaben sind durch eine Auflistung der vom Hersteller gewünschten technischen Details erläutert. Diese können von einer Kundenkonzeption bis zu einem bekannten, funktionierenden Bauelement reichen.

Die enthaltenen halb-spezifischen (semicustom) Bauteile beschränken sich auf Gatter- und Zellen-Arrays, die durch die letzte Maske zum kundenspezifischen Schaltkreis werden. Feldprogrammierbare Bauteile sind in diesem Abschnitt nicht enthalten. PALs werden unter Speichern (Memory) aufgeführt und PLAs finden sich sowohl im Memory- als auch im Digitalteil.

# INTRODUCTION AUX CIRCUITS FAITS SUR COMMANDE

Nouvelle cette année, cette Section décrit les services et produits fournis par des sociétés fabriquant sur mesure des circuits digitaux, linéaires, digitaux/linéaires. Le Guide Général de Sélection offre sur plusieurs pages un tableau indiquant par fabricant les services offerts, les équipements disponibles, les systèmes technologiques, et les procédés de test utilisés. Les possibilités de collaboration entre le fabricant et le client sont également indiquées dans ce même tableau.

Les produits partiellement faits sur mesure ont été limités aux "portes" et "cellules de mémoire" qui sont individualisées par un procédé final d'interconnexion. Les appareils programmables à l'extérieur ne sont pas considérés dans cette Section. Les PALs sont étudiés dans la Section "Mémoires", et les PLAs se trouvent dans les Sections "Mémoires" et "Systèmes Digitaux".

# INTRODUCCIÓN A LOS CIRCUITOS POR PEDIDO

Esta sección, nueva para este año, describe servicios y productos de compañías que ofrecen circuitos por pedido, sea digital, lineal y digital/lineal combinado. Apareciendo en orden alfabético en la Guía Maestra de Selección, las compañías se caracterizan por descripciones de servicio de diseño, taller de producción, tecnología de procesos y capacidad de distintas pruebas. Opciones de contribución por el operador en los proyectos de circuitos por pedido, están delineados en la lista de nivel de asistencia del operador preferido por el fabricante que abarca de concepto de sistema a pieza comprobada buena.

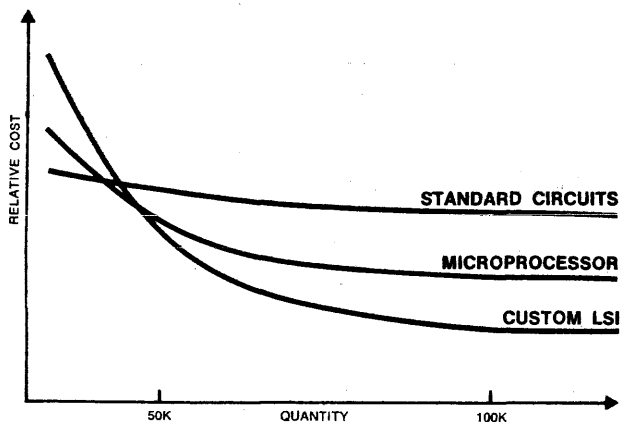
Componentes a pedido limitado están restringidos a redes de puertas y celdas electrónicas que son "a pedido" en el proceso final de interconexión. Componentes programables en el campo no están incluidas en esta sección. Los PALs aparecen en la lista la sección de Memoria y los PLAs se encuentran en ambas secciones bajo Memoria y Digital.

## カスタムサーキットへの案内

本年新たに加わったこのセクションにはデジタル、リニア、双方兼ねたカスタム回路を供給するメーカーのサービスと製品を掲載しています。メーカー名はABC順にマスターセクションガイドにのっていますが、そのメーカー毎に、デザイン、製造設備、プロセス技術、テスト能力を示してあります。カスタムサーキットに関するメーカー側のカスタマー参加希望はシステムの概念から良く知られている製品の良品迄概略を記載しています。

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# Options for Going Custom



Standard circuits are recommended for production volumes below 1000 units. Intermediate-volume applications may benefit best from customizing through software, such as with microprocessors. Custom chips have the lowest cost at high-volume production. Source: National Semiconductor

For many applications, standard integrated circuits may be inappropriate from the standpoints of cost, size, power consumption or reliability. Moreover, unique features demanded by proprietary products often require entirely new circuit configurations. As a result, customized ICs are assuming an increasingly important role in system design.

Custom IC suppliers report that the chief benefits enjoyed by nearly all custom-circuit users are low-cost parts and cost savings resulting from reduced printed-circuit board space, parts handling, inventory, testing requirements and system maintenance. Obtaining these benefits requires careful consideration of the many options provided by both custom ICs and other approaches which ultimately effect economics.

For example, in addition to standard and custom parts, options for implementing new system designs include semicustom ICs, microcomputers, custom microcomputers, a mixture of microcomputers and custom ICs, or a mixture of all of these. The system development strategy used depends largely upon marketing objectives and may require staged system development, first with standard ICs, then with semicustom ICs and, finally, full custom units. Or, the strategy may dictate developing standard or semicustom prototype systems with concurrent verification of a full custom design.

Another option is to alter a standard microprocessor and other standard circuits, rather than using a full custom design. Customizing standard products can reduce the design costs, turnaround time and risks of a full custom design. In some cases, semicustom or custom circuits can replace microprocessors which have been used in dedicated, mostly controller-type applications. Or, a custom circuit may be a direct integration of several standard ICs, such as op-amps, comparators and resistor networks.

All approaches require up-front decisions involving design, prototype and production turnaround times, volume/cost trade-offs, alternate sourcing, circuit configuration and process technologies, and the user/supplier interface. The most critical factor, however, is cost.

The cost of a nonstandard IC includes expenses for design and tooling, wafer and chip processing, packaging and testing.

**Design and Tooling:** Until recently, IC users had just two options for implementing new designs: standard parts and full custom circuits. When standard parts were inadequate, a user had to commit to great production volumes to amortize high development costs. Additionally, development times often extended well over a year, and chances of initial success were relatively slim. However, custom suppliers have minimized these drawbacks by devising new customizing techniques to the extent that few custom circuits now are developed entirely from scratch.

**Full Custom:** In this approach the circuit designer draws from a collection of time-tested circuit modules and components to customize a chip. These elements of known performance are located on the chip with the assistance of CAD equipment to form optimized interconnect patterns, thereby minimizing chip area and thus cost. Full custom design gives the most efficient use of silicon chip area. Although design turnaround time has been longer than other options, advanced CAD techniques are closing the gap.

**Semicustom:** This approach produces custom circuits by interconnecting repetitive patterns of preprocessed circuit elements on a chip called a *masterslice*. Because a masterslice is processed just short of the final interconnect pattern, the same part is mass-produced for use by all customers, with customizing occurring at the final interconnect stage.

Gate arrays are masterslices containing repetitive patterns of transistors connected as logic gates. Device arrays are patterns uncommitted transistors and resistors. The next level of customization uses an array of unconnected transistors and resistors called cells. Each cell can be interconnected internally to provide a specific logic function, and each cell on the chip interconnected into a customized system. The repertoire of allowable cell functions is called the cell library.

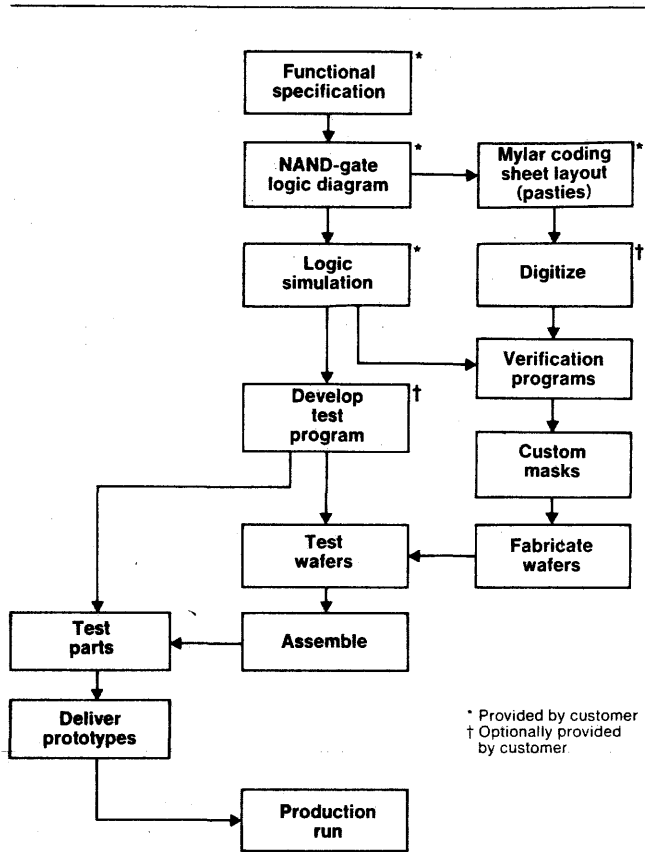
Interconnect design for masterslices can be manual or computer-aided. In all cases, the masterslice approach minimizes design turnaround time. Complex custom-cell library circuits can be obtained in about 18 weeks or less in prototype form; prototypes from gate arrays typically are available in about nine weeks.

One major disadvantage of masterslice circuits has been low circuit density. Circuit and device arrays require channels and alleys for routing interconnects. And because a single pattern must accommodate many different system designs, considerable interconnect routing space can be left unused after the chip is designed. Conversely, in a full custom design, unused space can be minimized. Consequently, a masterslice can be three times larger than an equivalent full-custom chip costing 20% to 60% less.

But masterslice manufacturers are increasing circuit densities to bring unit cost closer to that of full-custom circuits purchased in high volume. Circuits with two and three layers of interconnects are being developed to allow more efficient use of chip area. Triple interconnect-level masterslices are already used in a recent generation of IBM computers. So while full-custom designers are reducing design turnaround time, masterslice proponents are reducing unit cost.

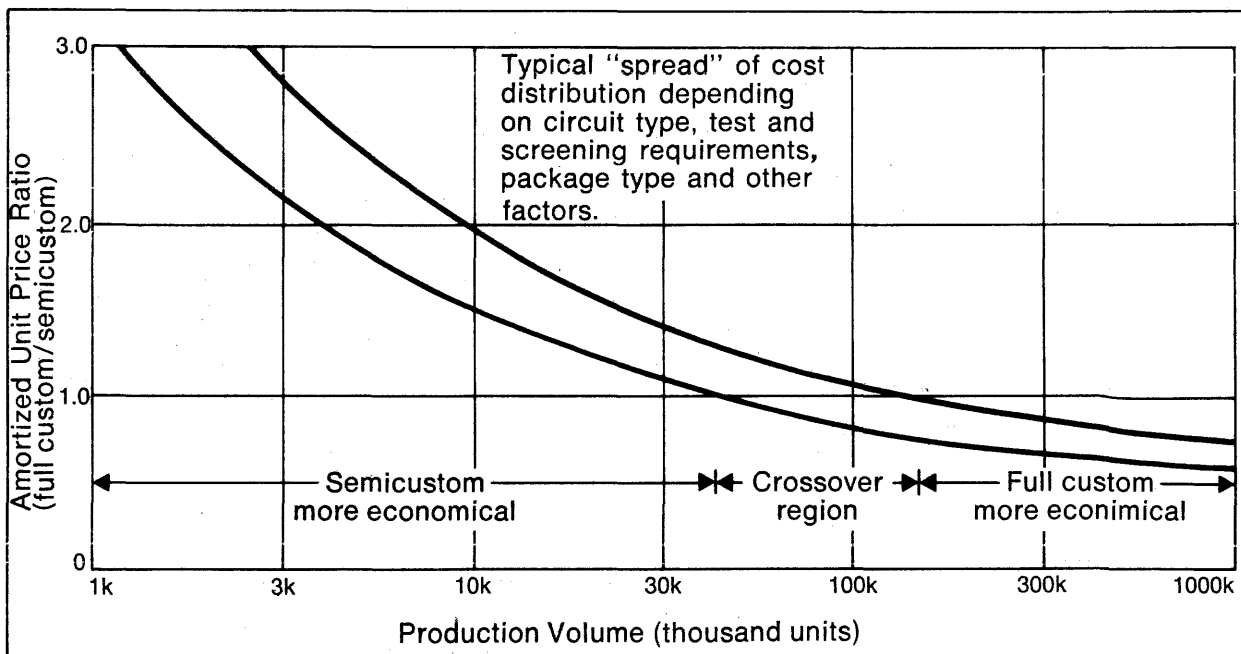
**Production:** IC houses that process custom circuits basically sell a production service, not a design service. Most are concerned first with recovering design costs and then making profit on the volume production of parts. The general rule of thumb calls for total production volume representing at least ten times the supplier's design cost.

Some firms may accept a smaller production volume if engineering costs can be lowered. The customer may satisfy this requirement by using a design specialty firm or the in-house



\* Provided by customer  
† Optionally provided by customer

Developing a custom IC is a multistage process. The degree of user involvement varies between companies. This diagram outlines the development of a custom chip designed from a cell library. Source: Signetics



Production volume influences the choice between full-custom and semicustom parts. However, other considerations (such as design turnaround time) may be overriding factors. Source: Exar

engineering staff. Specialty firms provide services such as design, mask-making, assembly and testing. In many cases a design house controls the evolution of an IC through tooling, fabrication and final delivery.

Selecting a fabrication process based solely on chip cost is not entirely straightforward. For any given process, all wafers cost about the same regardless of circuit complexity. The cost of a chip is determined chiefly by yield, which tends to decrease with chip-area increase. Thus, chip cost varies directly with chip area.

Cost comparison between processes is more complex. Older processes do not necessarily produce higher yields than newer process techniques. And although processes involving more fabrication steps tend to have higher production cost per wafer, a more complex process may also allow circuit designs that produce smaller chips. Similarly, wafers fabricated by newer processes that involve expensive production equipment are usually costly. But, improved processes, although expensive, can produce high yields which tend to lower the die cost. The process must therefore be chosen to get the lowest chip cost consistent with meeting performance requirements.

**Packaging:** A standard package can cost from a few cents to several dollars. In many custom designs, package cost can

be the largest part of unit cost, especially if a custom package is required.

**Testing:** This cost can be a large portion of the total unit cost. So care must be taken to design the chip for minimum testing requirements. Testing costs depend on the type of test equipment used and the amount of time required to test. Ideally the simplest and most widely available equipment should be used.

Testing cost can range from \$50 per hour to several hundred dollars per hour. The manner in which tests are specified can vary test time per part from a few seconds to a minute. It is not economically feasible to test many complex digital circuits for every possible combination of inputs and outputs. And unless complex random logic is designed with testing in mind, it may not be possible to test the final product adequately.

Testing should be considered early in the design phase to minimize cost. Usually it is possible to select a process technology or circuit configuration that eliminates the need for some parametric tests. If timing requirements are critical, for example, using one of the faster process technologies or circuit configurations can eliminate testing speed parameters. Similarly, using a superior process for linear circuits could eliminate the need to test for offset voltages in op-amps.

## Terminology

**Assembly:** Process of mounting a chip into a package and connecting the chip terminal pads to package terminals.

**CAD:** Computer-aided design. CAD includes computerized equipment that performs circuit simulation, logic simulation, automatic circuit and logic drawing, topological digitizing, topology construction on a CRT terminal, design rule checking and test generation.

**Cell:** Circuit performing a digital or linear function that is repeatedly used to design an LSI/VLSI chip.

**Cell library:** Collection of predesigned cell functions stored in a CAD data base. Custom LSI/VLSI devices can be designed by choosing appropriate cells from the library and locating them on a chip to minimize interconnects and maximize performance. The cells are computer characterized for performance much like SSI and MSI ICs with data sheets for each cell.

**Custom circuit:** In general, a component whose manufacture is under the exclusive control of a customer. The term can refer to full-custom, customized-standard or semicustom parts. In semicustom parts most of the mask layers are common to many customers, and only the final interconnect patterns are special. A customized part is a modification of a standard part to the requirements of a customer. Full-custom parts normally are fabricated from masks configured for the customer.

**Design rules:** Collection of rules that define minimum dimensions of device topological structures. Design rules also express process-parameter design limits such as gain factor, threshold level, oxide thickness and capacitance.

**Die:** Rectangular piece of semiconductor material into which electrical circuits have been fabricated. Also called a chip. Plural is dice.

**Digitized data base:** Recorded digital data representing a topological drawing of an SSI, MSI, LSI or VLSI device. The data include locations and dimensions of rectangles that make up individual circuit elements and interconnects.

**Feature size:** Dimensions of rectangles, lines and spacings in an IC topological design.

**Gate:** Basic digital-logic element producing a binary output depending on the logic state of various inputs.

**Gate array:** Regular pattern of circuit components on a chip, connected to form regular patterns of gates. The gates are not interconnected until the customer specifies the chip function.

**Gate equivalent:** Basic unit of measure for digital circuit complexity based on the number of elementary logic gates needed to provide the same circuit function.

**LSI:** Large-scale integration. Device design integrating from 100 up to thousands of gate equivalents on a chip.

**Macrocell array:** Regular pattern of grouped, unconnected circuit components. Macrocells are formed into standard logic elements by interconnecting the components to provide specific circuit functions. Interconnects between the cells are specified to perform a specific chip function. The formation of the cell functions and cell interconnects are unique to the customer.

**MSI:** Medium-scale integration. Device design integrating from 10 to 100 gate equivalents on a chip.

**Masterslice:** In general, a partially processed chip containing circuit elements for customizing through final metal interconnect patterns. Can refer to gate, device or cell arrays.

**Pastie:** Scaled decal (usually transparent) representing both function and dimensions of an IC building block such as a gate, flip-flop or I/O buffer. Pasties are a commonly used tool for designing LSI/VLSI chips from a cell library. Each cell type has pastie equivalents.

**SSI:** Small-scale integration. ICs containing fewer than ten logic-gate equivalents.

**Silicon gate:** MOS design in which the gate is made of silicon instead of metal. Silicon-gate MOS is faster and more dense than metal-gate MOS.

**VLSI:** Very-large-scale integration. Device design integrating thousands of gate equivalents on a chip.



**Cost:** Intense activity in both gate arrays and associated development tools should make the gate-array concept extremely cost competitive with other design approaches. Typical costs for the development of a gate array and a full-custom IC are shown in the accompanying tables.

New gate arrays are currently in widespread development by IC manufacturers; CMOS products with up to 8,000 gates and emitter-coupled-logic (ECL) devices with 100-pico-second gate delays have already been developed. Although most of the attention has been concentrated on CMOS, many bipolar technologies are not being neglected.

Gate arrays are available in a number of bipolar technologies including ECL, Schottky TTL, integrated injection logic, integrated Schottky logic (ISL), and Schottky transistor logic (STL). Gate delays for these devices can be as low as 0.8 nanoseconds. Gate array chips are also available in NMOS and combination CMOS/NMOS versions.

Availability of computer-aided-design (CAD) tools and software support are helping both semicustom and custom products gain acceptance. In some cases, IC suppliers permit customers to use the supplier's in-house CAD facilities. Training courses, running three to five days, are also being offered by many IC manufacturers.

#### Custom IC cost factors.

##### Before Process Assembly

- Orders: Issues & Tracking
- Specific Documentation
- Orders: Issues & Tracking
- Purchase Prices
- Cost of Capital
- Inventory: Storage Space/Handling
- Incoming Inspection
- PC Board Space
- Power Consumption

##### After PC Assembly

- Test: Board
- Failure Diagnostic
- Test Active/Passive Components
- Repair
- Re-Test
- Re-Inventory

#### Typical costs for the development of a 1000-gate array.

	min.	max.
<b>1. Tooling Development</b>		
A) Circuit Conversion	\$500	\$1500
B) Layout	\$2.5K	\$5K
C) Digitization, PG Tape, Masks, 20 prototype units in ceramic DIP, bench tested	\$9K	\$13K
<b>2. Preproduction</b>		
A) Test Program Development	\$4K	\$8K
B) Test hardware		
Personality Board	\$1K	\$2K
Probe Card	\$100	\$200
Special Supplies, Signal Generator	*	*
Burn-In Board	\$500	\$2500
Other	*	*
C) Production Qualification Units (Optional)		
Burned-In, Temperature Tested 100 units	3K	\$6K
883B Environmental Screening	Extra	\$40K
TOTAL	\$18K	\$40K
* Depends on application		

#### Typical development costs for a full-custom integrated circuit.

	min.	max.	
<b>1. Tooling Development</b>			
A) Circuit to LSI Conversion			
B) Layout			
C) Digitization, PG Tape, Masks	\$20K	\$200K	
<b>2. Engineering Evaluations: 5 wafers</b>			
Working Plates	Per Plate	\$50	\$100
Bipolar: 1-Layer Metal	4-6 weeks	\$3K	\$6K
2-Layer Metal	6-8 weeks	\$5K	\$10K
Pt Schottky Diodes	- week extra		
CMOS: 1 Poly, 1 Metal	6-8 weeks	\$3K	\$5K
Same, 4 Micron	6-8 weeks	\$4K	\$6K
May need several iterations, usually 2 to 4			
<b>3. Prototypes</b>			
A) Establish Waferbank			
30-45 Wafers from Several Runs	\$10K	\$30K	
B) First Look Samples			
100 Untested Dice Packaged in Ceramic Sidebraze	\$600	\$1500	
C) Test Program Development	\$5K	\$30K	
D) Test Hardware			
Personality Board	\$1K	\$2K	
Probe Card	\$100	\$200	
Special Supplies, Signal Generators	*	*	
Burn-In Board	\$500	\$2500	
Other	*	*	
E) Production Qualification Units (if desired)			
Burned-In, Temperature Tested 100 Units	\$3K	\$6K	
883B Environmental Screening	Extra	\$280K	
TOTAL	\$40K	\$280K	
* Depends on application			

**Comparison.** The relative ranking of risk factors for alternative design approaches is shown at the right. These ratings will vary depending on the selected suppliers for custom, cell library, gate array or non-LSI devices. Therefore, this chart should be used only as a guide. In the case of linear designs, risks are greater than for digital counterparts. Unless the various circuit blocks assembled from a cell library are compatible with each other from the standpoint of processing, the overall performance of the chip can be less than expected. For instance, the breakdown voltage, beta, sheet resistance, implant dosage, and epi thickness requirements, to say nothing of the starting material, may be mutually exclusive. Also, NPN and PNP transistor configurations cannot both be optimized for performance as is possible with discrete devices.

**Comparisons of advantages of various design approaches. A rating of 1 indicates best. (Numbers in parenthesis apply to linear designs.)**

	Full Custom	Cell Library	Gate Array	Non-LSI
Design Costs	4(3)	3(4)	2	1
Design Time	4(3)	2(4)	3(2)	1
Mask Costs	4	4	2	1
Redesign Flexibility	4	3	2	1
Test Program Costs	3	3	2	1
Circuit Purchase Price	1	2	3	4
System Power Required	1	2	2	4
Reliability	1	2	2	4
PC Board & Costs	1	2	2	4
Production Labor	1	2	2	4
Security	1	3	2	4
Added Features/Board	1	3	3	4

## CUSTOM CONSIDERATIONS

Benefits provided by custom, related to cost factors, are described in the following paragraphs:

**Specification Documentation:** Since a single custom IC can replace as many as 100 MSI circuits (plus assorted external active and passive components, such as decoupling capacitors, diodes, and transistors), paperwork needed is greatly reduced.

**Purchasing:** The purchasing function is not a "free" activity in any company. It costs money to issue purchase orders, and to track them through various delivery dates and procedures with phone calls and computer time. The lower parts count significantly reduce this paper load.

**Purchase Price:** Depending on the complexity of the chip, the price for a custom device can be lower, at, or above that of all the components to be replaced. It is important to count all of the passive peripheral components involved, such as sockets, resistors, capacitors, inductors, and perhaps even connector pins to the outside.

**Cost of Capital:** This cost varies with prevailing interest rate. If a company has to borrow development money, usually at about three or four percent over prime rate, then this interest is also an expense incurred by the project. But even if cash is available, there still may be an opportunity cost. This is the income that could have been earned if the cash had been used on

short-term projects (such as buying more inventory — provided the sales to turn the inventory over exists).

**Inventory:** Once the material is in the plant, it has to be handled (including counting, sorting, and paperwork), and stored. Usually, storage space is predefined, and, since it already exists, it is not considered to be an additional expense. However, one has to remember that real estate value on a per-footage basis is substantial and that other departments might make a more cost effective use of any space available. Hence, an additional cost is incurred for each additional component that has to be stored and accounted for. Also, components may become obsolete before the inventory is used up. Custom ICs contribute to decreasing inventory expenses.

**Incoming Inspection:** Unless pre-aged and pre-screened components are bought to prevent early failures (infant mortality), it is advantageous to inspect active components as they come in. Again, a reduction in the number and variety of devices to be tested adds to profits.

**Printed Circuit Board Space:** Fewer components mean less space needed to mount them, less auxiliary components, less artwork for interconnect lines, fewer holes to be drilled, and less insertion time and effort — as well as a smaller board.

**Power Consumption:** The power consumption of the entire system is usually dramatically reduced, making savings possible with lowered power supply and cooling requirements.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM

Manufacturer	Advanced Micro Devices (Page 4600)	Alphatron	American Microsystems, Inc. (Page 4614)
FOR DETAILED DATA SEE:			
Customized Standard Circuits			Digital, Linear, Combined Digital/Linear
Gate Array	Programmable Array Logic		3 $\mu$ and 5 $\mu$ silicon-gate CMOS
Chip Density Range (equiv. gates)	200 to 800	Up to 4400 2-input gates	300 to 1200 2-input gates (5 $\mu$ ) 500 to 5000 2-input gates (3 $\mu$ )
Cell Library		CMOS	Yes
Design Kit Available	Yes	Yes	Yes
Full Custom Circuits			
Digital		Silicon- or metal-gate PMOS, NMOS, CMOS	Silicon- or metal-gate PMOS, NMOS, CMOS with high-voltage output capability.
Linear		Silicon- or metal-gate PMOS, NMOS, CMOS	Silicon-gate NMOS, CMOS, filter, amplifier to rf
Combined Digital/Linear		Silicon- or metal-gate PMOS, NMOS, CMOS	Silicon-gate NMOS, CMOS, 14-bit resolution, VHF-rf
Provide Design Assistance	Yes	Yes	Yes - class instruction
Acceptable Customer Input (in order of preference)		Alphamap layout logic diagram functional description block diagram.	Logic diagram, functional specification, data-base tape, PG tape, masks, COT.
Design Aids	AMPALASM-20 software for Boolean equation generation. AMPLPL software provides similar capability for advanced products.	Logic simulation, breadboard assistance, design rule checks, decals.	Complete CAD assistance for logic circuit simulation design rule checks, breadboard assistance.
Production	In-house	Wafer production procured from outside foundries. Testing done in-house.	Masks, fabrication and assembly in-house.
Preferred Delivered Product	Ceramic and plastic DIPs, leadless chip carriers, dice; all available in commercial and military.	Any upon request	DIP, chip carriers, die-on-board packages, wafers, die to commercial or military specifications.
Test Program Generation		Yes	Yes, CAD supported
Production Test	100% dc, ac and functional testing. Burn-in, thermal shock, environmental, MIL.	Functional, parametric, burn-in, thermal shock, environmental, MIL.	Functional, Parametric, burn-in, thermal shock, environmental, MIL.
Electrical Test Systems Available	Xincom, Accutest	In-house designed, customer-supplied, or outside service.	Sentry, Teradyne, Xincom, LTX, General Radio.
Comments	Commercial programmers available Software output JEDEC PLDTF. Most complex programmable logic parts available.	Multiple sourcing. Cell library and design rule handbook.	Complete custom capability including design, instruction, CAD and process licensing and technology sales. CAD also available through timesharing and at Design Centers.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Applied Micro Circuits	AWI	Barvon Research (Page 4626)
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>		Linear, Digital, Combined Digital/Linear	Digital Combined Digital/Linear
Gate Array	ECL	Bipolar, CMDS, silicon-gate MOS, hybrids	2.5 $\mu$ silicon-gate CMOS, Bi-CMOS (Bipolar and CMOS)
Chip Density Range (equiv. gates)	250 to 1500 gates	10 to 10,000	100 to 10,000 2-input gates
Cell Library	ECL	Over 2600 type	CMOS, Bi-CMOS
Design Kit Available	Yes	Yes	Yes
<b>Full Custom Circuits</b>			
Digital	ECL	Yes	Silicon-gate CMOS, HCMOS, HMOS, Bi-CMOS
Linear		All processes	Silicon-gate CMOS, HCMOS, Bi-CMOS
Combined Digital/Linear		All processes	Silicon-gate CMOS, HCMOS, Bi-CMOS
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Functional specification, logic diagram, test vectors, PG Tape, Tegas Tape.	Functional schematic, circuit diagram	Logic diagram, circuit diagram, breadboard, functional speci- fication, test vectors, COT, known good device.
<b>Design Aids</b>	Logic simulation, design rule checks.	Users manual	Circuit simulation, logic simula- tion, schematic entry, test program generation, design rule checks, breadboard assistance.
<b>Production</b>	In-house, procured	In-house robotics and automation	In-house, procured
<b>Preferred Delivered Product</b>	Packaged dice	Printed card mounted, packaged units	Packaged devices, others available
<b>Test Program Generation</b>		No	Yes
<b>Production Test</b>	Functional, parametric, burn-in, thermal shock environmental.	Functional	Functional, parametric, burn-in, thermal shock, environmental, MIL.
<b>Electrical Test Systems Available</b>		Customer supplied	Sentry, Teradyne
<b>Comments</b>			Complete custom capability includ- ing design, instruction, CAD and technology licensing. Standard cell library available.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	California Devices	Cherry Semiconductor	Circuit Technology Inc.
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital	Digital, Linear, Combined Digital/Linear	Digital
Gate Array	Silicon- or metal-gate CMOS	1 <sup>2</sup> L, LSI <sup>2</sup> L, ECL and other bipolar	Silicon gate CMOS
Chip Density Range (equiv. gates)	50 to 1780 2-input gates	64 to 150 gates/mm <sup>2</sup>	560 to 1440 — 5 micron single metal 1120 to 3876 — 4 micron double metal
Cell Library	Silicon- or metal-gate CMOS	1 <sup>2</sup> L, LSI <sup>2</sup> L, ECL and other bipolar	CMOS and LSTTL
Design Kit Available	Yes	Yes	Yes plus CAD
<b>Full Custom Circuits</b>			
Digital		1 <sup>2</sup> L, LSI <sup>2</sup> L, ECL and other bipolar	Silicon gate CMOS
Linear	Bipolar linear arrays	1 <sup>2</sup> L, LSI <sup>2</sup> L, ECL and other bipolar	
Combined Digital/Linear		1 <sup>2</sup> L, LSI <sup>2</sup> L, ECL and other bipolar	
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Logic diagram with test vectors and PG tape; circuit diagram with data-base tape and composite drawing; breadboard.	Customer-owned tooling; circuit diagram; logic diagram; breadboard.	Logic diagram and test vectors. Customer-owned tooling. Functional specification.
<b>Design Aids</b>	Logic and circuit simulation; breadboard assistance; design rule checks.	Logic simulation; breadboard assistance; design rule checks.	Logic simulation, auto layout, rules checks, network comparison checks.
<b>Production</b>	In-house and procured	In-house	In-house plus second sourcing availability.
<b>Preferred Delivered Product</b>	Scribed dice, packaged dice, substrate-mounted device.	Packaged dice; flip chips	All options available
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	Functional, parametric, burn-in, thermal shock, environmental, MIL.	All except military	Full capability including full MIL, class S and geothermal (200°C).
<b>Electrical Test Systems Available</b>	Sentry VII, Sentry Series 20	Teradyne 1259, A310, A311	Fairchild series 10
<b>Comments</b>	Si-gate isolated oxide CMOS has TTL/LSTTL interfacing capability, high-speed performance.	Dice can be solder-bump flip chips with nitride passivation; packaged dice can be delivered from COT for full custom.	No limit on package variety. In-house multichip hybrid capability.

CUSTOM/SEMICUSTOM (cont)

Manufacturer	Circuit Technology Inc.	Citel	Comlinear Corp.
<b>FOR DETAILED DATA SEE:</b>		(Page 4628)	
<b>Customized Standard Circuits</b>	Digital	Digital	Yes
Gate Array	CMOS on Sapphire	3 $\mu$ silicon-gate CMOS	
Chip Density Range (equiv. gates)	1120 to 3876-4 micron double metal	200 to 1200	
Cell Library	CMOS and LSTTL	Yes	
Design Kit Available	Yes plus CAD		
<b>Full Custom Circuits</b>			
Digital	CMOS on Sapphire	Silicon or metal-gate CMOS, silicon-gate NMOS, bipolar	
Linear		Bipolar	Bipolar-FET
Combined Digital/Linear		CMOS	
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Logic diagram and test vectors. Customer-owned tooling. Functional specification.	Logic diagram, circuit diagram, functional specification, bread-board, data-base tapes, PG tapes, mask sets, customer-owned tooling.	Functional specification
Design Aids	Logic simulation, auto layout, rules checks, network comparison checks.	Logic simulation, design-rule checks, net-list verification.	Computer simulation
Production	In-house	In-house	In-house or procured
Preferred Delivered Product	All options available	Packaged units; will supply wafers or dice.	Packaged circuit, printed card mounted units.
Test Program Generation	Yes	Yes	Yes
Production Test	Full capability including full MIL and class S.	Functional, parametric	Functional, burn-in, thermal shock, linear, MIL
Electrical Test Systems Available	Fairchild series 10	Sentry VII	Complete time domain and frequency domain testing from dc to 18 GHz.
Comments	No limit on package variety. In-house multichip hybrid capability.	Complete custom capability	Specialties include amplifiers with extremely wide bandwidth and fast settling time, fast sample-hold and A to D conversion products.



# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Custom Integrated Circuits	Custom MOS Arrays (Page 4630)	Exar Integrated Systems (Page 3386)
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital, Linear, Combined Digital/Linear	CMOS	Digital, Linear, Combined Digital/Linear
Gate Array	I <sup>2</sup> L, Linear Master Slice, design on other companies' gate arrays		Metal-gate CMOS; I <sup>2</sup> L Si-gate CMOS
Chip Density Range (equiv. gates)	50 to 8000 (5-input gates)	448 to 1568 (2 inputs)	100 to 1250
Cell Library	I <sup>2</sup> L	Yes	Metal-gate CMOS; I <sup>2</sup> L Si-gate CMOS
Design Kit Available			Yes
<b>Full Custom Circuits</b>			
Digital	I <sup>2</sup> L, CMOS, TTL and other bipolar devices		Metal-gate CMOS; I <sup>2</sup> L Si-gate CMOS
Linear	Bipolar, CMOS		No
Combined Digital/Linear	I <sup>2</sup> L/Linear, TTL/Linear, CMOS/Linear		I <sup>2</sup> L, other bipolar
Provide Design Assistance	Yes		Yes
Acceptable Customer Input (in order of preference)	No preference		Complete layout, breadboard, circuit diagram, logic diagram, functional diagram.
<b>Design Aids</b>	Logic simulation, circuit simula- tion, design rule checks, breadboarding, complete Calma layout system, in-house CAD software, engineering assistance.		Logic simulation; breadboard assistance; design rule checks.
<b>Production</b>	Procured		In-house
<b>Preferred Delivered Product</b>	No preference		Probed wafers, scribed dice, packaged dice.
<b>Test Program Generation</b>	Yes		Yes
<b>Production Test</b>	All		Functional, parametric, environmental, burn-in
<b>Electrical Test Systems Available</b>	HP, MCC		Sentry; Fairchild 5000, Teradyne J273, A311; Amdahl
<b>Comments</b>	Total IC development from con- cept through product delivery. Any subset of the development cycle (design, layout, etc.)		

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Fairchild	Ferranti	Fujitsu
<b>FOR DETAILED DATA SEE:</b>	(Page 4634)		(Page 4637)
<b>Customized Standard Circuits</b>	Digital	Linear, Digital, Combined Linear/Digital	Digital
Gate Array	Bipolar ECL	Collector Diffusion Isolation	Silicon-gate CMOS, LSTTL
Chip Density Range (equiv. gates)	20 to 1000 3-Input NAND	200 to 10,000 Gates/chip	200 to 8000 2-Input NAND
Cell Library		Collector Diffusion Isolation	Silicon-gate CMOS, LSTTL
Design Kit Available		Yes	Yes
<b>Full Custom Circuits</b>			
Digital		Collector Diffusion Isolation	Silicon-gate CMOS, LSTTL
Linear		Collector Diffusion Isolation	Silicon-gate CMOS, LSTTL
Combined Digital/Linear		Collector Diffusion Isolation	
Provide Design Assistance		Yes	Yes
Acceptable Customer Input (in order of preference)		Pattern-generator tape, composite drawing when done on Ferranti's supplied design station; functional specification, logic diagram, circuit diagram, breadboard, test vectors.	Logic data and test timing data (Fujitsu format).
<b>Design Aids</b>	Logic simulation, breadboard assistance, design rule checks.	Logic simulation, breadboard assistance, design rule checks, full design capability.	Logic simulation, design rule checks; complete CAD system.
<b>Production</b>	In-house	In-house	In-house
<b>Preferred Delivered Product</b>		Packaged dice	Substrate mounted dice
<b>Test Program Generation</b>		Yes	Automatic
<b>Production Test</b>		Functional, parametric, burn-in, thermal shock, environmental, MIL, linear.	Functional, ac and dc parametric.
<b>Electrical Test Systems Available</b>	FARCAD	LTX, Teradyne, Custom	In-house
<b>Comments</b>	Remote access to FARCAD via telephone link	Semi-custom from off-the-shelf selection of a range of standard wafers.	Dual-layer metal; CAD system verifies design; auto placement and routing from customer's inputs.



# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	General Instrument	GTE Microcircuits	Harris Semiconductor
<b>FOR DETAILED DATA SEE:</b>			(Page 4644)
<b>Customized Standard Circuits</b>			Digital, Linear, Combined Digital/Linear
Gate Array	Silicon-gate CMOS	ISO-CMOS	Silicon-gate CMOS, TTL-AS, TTL-ALS
Chip Density Range (equiv. gates)	560 to 2014	504 — 2500	Up to 10,000 (CMOS) Up to 2,500 (TTL-AS) 120, 312 and 918, (TTL-ALS) (Rad Hard)
Cell Library	Silicon-gate CMOS	ISO-CMOS	Silicon-gate CMOS, TTL-AS
Design Kit Available	Yes	Yes	Yes
<b>Full Custom Circuits</b>			
Digital	No	ISO-CMOS	Metal-gate CMOS and MOS; silicon-gate PMOS, LSTTL; ALSTTL; STTL; I <sup>2</sup> L.
Linear	No	ISO-CMOS	Bipolar (dielectric isolation) CMOS/bipolar (metal-gate/ bipolar) and silicon.
Combined Digital/Linear	No	ISO-CMOS	Silicon-gate CMOS; bipolar
Provide Design Assistance	No	Yes	Yes
Acceptable Customer Input (in order of preference)	No	Calma tapes, PG tapes, masks	System specification, Logic diagram, JG, CALMA or Mebes Data Base.
Design Aids	No	Design rule checks, Logic simulation	Proven Cell Library MACROS, Logic Simulation, Layout Pro- grams and Test Grading.
Production	No	In-house	In-house
Preferred Delivered Product	No	Packaged die, probed wafers, mapped wafers	Wafer, packaged dice, scribed dice, and packaged devices.
Test Program Generation	No	Yes	Yes
Production Test	No	Functional, parametric, burn-in	Up to and included equivalent Class "S" military flow. Also in-house total dose testing.
Electrical Test Systems Available	No	Sentry LTX, H.P., Mega, XINCOM	Sentry
Comments			Also offer low voltage CMOS. Radiation hardening for most process technologies. Also provides silicon foundry.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Holt, Inc.	Integrated Circuit Systems	Integrated Technology Corporation
<b>FOR DETAILED DATA SEE:</b>	(Page 4745)		
<b>Customized Standard Circuits</b>	Linear, Digital, Combined Linear/Digital	Digital, Linear, Combined Digital/Linear	
Gate Array	Metal gate CMOS, linear bipolar	CMOS	
Chip Density Range (equiv. gates)	80 to 1098		
Cell Library	Metal gate CMOS	All MOS processes	
Design Kit Available	No		
<b>Full Custom Circuits</b> Digital	Silicon gate, metal gate CMOS	CMOS/SOS; CMOS, DMOS, NMOS, SI or Metal Gate; Double level metal; Double level poly.	Silicon- or metal-gate MOS; TTL, STTL, LSTTL, I <sup>2</sup> L, ECL, other bipolar
Linear	Bipolar	CMOS	Silicon- or metal-gate CMOS; bipolar
Combined Digital/Linear	Metal gate CMOS and bipolar	CMOS	Silicon- or metal-gate CMOS; bipolar
Provide Design Assistance	Yes	Yes	
Acceptable Customer Input (in order of preference)	Any	No minimum required	
<b>Design Aids</b>	Logic simulation, breadboard assistance, DRC.	DRC and ERC on VAX 11/780, Calma GDS II, sticks software, production test, systems.	Logic simulation; breadboard assistance; design rule checks; transient analysis.
<b>Production</b>	Procured	Small to medium quantity	Procured
<b>Preferred Delivered Product</b>	Packaged die, scribed die	Packaged parts	Mapped wafers; probed wafers; packaged dice; PC-mounted packaged units; design tooling; complete systems.
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	Functional, parametric, burn-in, thermal shock, environmental, MIL, Linear SEM.	Yes	Functional; parametric
<b>Electrical Test Systems Available</b>	IT-200	Megatest	Siemens
<b>Comments</b>		Services: design, layout, chip/PCB artwork, test generation, product engineering, test and production.	Services: design, tooling for custom ICs; service industry used for photomasks, wafer fab, assembly, some testing. System design and production with custom IC's.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Interdesign (Page 4747)	International Microcircuits (Page 4748)	International Microelectronic Products (Page 4749)
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital, Linear, Combined Digital/Linear	Digital	Digital, Combined Linear/Digital
Gate Array	CDI RTL and CML, metal-gate, CMOS, silicon-gate CMOS; bipolar	Silicon- or metal-gate CMOS	No
Chip Density Range (equiv. gates)	See comments	75 to 7500	200 to 10,000
Cell Library	Yes	Yes	Silicon-gate CMOS
Design Kit Available	Yes	Yes	Yes
<b>Full Custom Circuits</b>			
Digital	Silicon-gate CMOS; metal-gate CMOS; CDI bipolar; Linear bipolar; Linear CMOS.		Silicon-gate NMOS and CMOS
Linear	Bipolar		
Combined Digital/Linear	All linear functions		Silicon-gate NMOS and CMOS
Provide Design Assistance	Yes		Yes
Acceptable Customer Input (in order of preference)	Pencil-connected layout; logic diagram; specification; PG tapes; circuit diagram and breadboard (for linear bipolar).		Functional specification, logic diagram, circuit diagram, breadboard, test vectors, customer-owned tooling (pattern-generator tape, composite drawing), known good device.
Design Aids	Logic/circuit computer simulation; breadboard parts; layout sheets; functional overlays; evaluation parts.		Logic simulation, breadboard assistance, design rule checks, engineering assistance.
Production	Metallized wafers procured; other functions in-house.	In-house and procured	In-house
Preferred Delivered Product	Packaged dice, bare dice, probed wafers.	Scribed dice; packaged dice	Mapped wafers, probed wafers, scribed dice, packaged dice, or special packages.
Test Program Generation	Yes	Yes	Yes
Production Test	Functional; parametric; burn-in; MIL 883B; environmental.	Functional, parametric, burn-in, thermal shock, environmental, MIL.	Functional, parametric, burn-in
Electrical Test Systems Available	Teradyne J325; LTX CP/TS70; LOMAC LM325; Fairchild 5000; LTX DX90, Pragmatic Inspector 100.	GenRad, Sentry	Sentry 20, Kiethley
Comments	CDI RTL; 255 logic cells; CDI CML 450-2000 gates; metal-gate CMOS: 140-800 gates. Silicon-gate CMOS 150-2000 gates. Linear bipolar 100-150 components.		



CUSTOM/SEMICUSTOM (cont)

Manufacturer	Intersil	LSI Computer Systems	LSI Logic
FOR DETAILED DATA SEE:	(Page 4751)	(Page 4761)	(Page 4762)
Customized Standard Circuits	Digital, Linear/Digital		Digital
Gate Array	Silicon-gate CMOS, 2 $\mu$ and 4 $\mu$		Silicon-gate CMOS, HCMOS and ECL
Chip Density Range (equiv. gates)	To 15,000		300 to 10,000
Cell Library	Yes		7400/4000 CMOS and HCMOS, ECL; 10K-SSI/MSI functions
Design Kit Available	Yes		Yes
Full Custom Circuits Digital	CMOS	Metal-gate PMOS, CMOS; Silicon-gate NMOS; Silicon-gate CMOS.	Implemented in gate array format Silicon gate oxide isolated CMOS/HCMOS; oxide isolated bipolar ECL.
Linear	CMOS	Metal-gate PMOS, CMOS; Silicon-gate CMOS	
Combined Digital/Linear	CMOS	Metal-gate PMOS, CMOS; Silicon-gate CMOS	
Provide Design Assistance	Yes	Yes	
Acceptable Customer Input (in order of preference)	Schematic, logic diagram, test vectors.	Logic diagram; customer-owned tooling (pattern-generation tape, composite drawing); circuit diagram; breadboard; functional specification.	Logic diagram to PG tape — Many input formats possible depending on customer interface required.
Design Aids	CAD assistance, training, manuals.	Design rule checks, computer-aided transient analysis, Applicon 760.	LSI design system (LDS), including design entry, circuit simulation, logic simulation, PG and test tape generation via remote terminal or factory based.
Production	In-house	Procured	In-house
Preferred Delivered Product	Packaged dice	Packaged dice, dice	Packaged dice
Test Program Generation	Yes	Yes	Yes
Production Test	Functional, parametric, burn-in, thermal shock, environmental, MIL.	Functional, parametric, burn-in; thermal shock, environmental; MIL.	Functional parametric to MIL 883B.
Electrical Test Systems Available	Sentry, Teradyne, Xicom, Tektronix	Macrodata 107 and customized equipment.	Industry standard tester
Comments	Specialize in linear and data-acquisition signal-conditioning circuits.	Multiple-sourced production	Multiple sourced

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Master Logic	MCE Semiconductor	Micrel Inc.
<b>FOR DETAILED DATA SEE:</b>			<b>(Page 4765)</b>
<b>Customized Standard Circuits</b>	Digital	Linear, Digital, Combined Linear/Digital, Bipolar, CMOS	
Gate Array	Silicon- or metal-gate CMOS	CMOS, ECL, Dielectric Isolation, I <sup>2</sup> L, Linear	
Chip Density Range (equiv. gates)	50 to 1500, silicon-gate 50 to 600, metal-gate	50 to 10,000 gates	
Cell Library	Yes	Yes	
Design Kit Available	Yes	Yes	
<b>Full Custom Circuits</b>			
Digital	Silicon- or metal-gate CMOS	CMOS, TTL, LSTTL, I <sup>2</sup> L, ECL, Linear	PMOS, CMOS, NMOS silicon or metal-gate, isoplanar type processing.
Linear	No	Up to 75 Volt	Bipolar, CMOS
Combined Digital/Linear	Yes	Up to 20 Volt	
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Any	MCE will interface with customer anywhere in design sequence. UniDES; software.	Customer owned tooling, PG Tape, Data base tape.
<b>Design Aids</b>	Design Manual, CAD with VIA Systems, "Gates" design package.	Logic simulation; breadboard assistance; design rule checks; UniDES.	Design rule checks
<b>Production</b>	Procured	In-house manufacturing; 4" Wafer Fab	In-house wafer fabrication; 3- and 4-inch lines procured assembly.
<b>Preferred Delivered Product</b>	Any	Mapped wafers; probed wafers; scribed dice; substrate-mounted dice; packaged dice; custom packaging.	Packaged tested die, probed wafers, mapped wafers.
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	In-house	Functional; parametric	Functional, parametric, burn-in, MIL STD 883.
<b>Electrical Test Systems Available</b>	Pragmatic designs	Teradyn J273 and A300 with laser trim; Pragmatic Inspector 200 (72 pin digital IC); Kiethley 300.	Sentry VII, Series 20
<b>Comments</b>	Multiple sourced	Functional arrays available CMOS MGA and SGA series; also linear function cells.	

**CUSTOM/SEMICUSTOM (cont)**

Manufacturer	Microcircuits Technology, Inc.	Micro Innovators	Micro Power Systems (Page 4766)
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital, Linear, Digital/Linear	Digital	
Gate Array	Linear bipolar, metal gate NMOS, CMOS; silicon gate CMOS		
Chip Density Range (equiv. gates)	250 to 6000 gates	50 to 250 mil <sup>2</sup>	
Cell Library	Silicon gate or metal gate NMOS, CMOS, linear bipolar		Poly-gate CMOS, bipolar LSI, thin film on silicon
Design Kit Available	No	No	No
<b>Full Custom Circuits</b>			
Digital	Silicon gate CMOS, Metal Gate CMOS	Silicon- or metal-gate PMOS, NMOS, CMOS, other MOS	Low-voltage, poly-gate CMOS
Linear	Bipolar, CMOS	Silicon- or metal-gate PMOS, NMOS, CMOS, other MOS	High-gain/ low current analog bipolar; thin-film resistor
Combined Digital/Linear	CMOS and bipolar	Silicon- or metal-gate PMOS, NMOS, CMOS, other MOS	Bipolar and CMOS
Provide Design Assistance	Yes		Yes
Acceptable Customer Input (in order of preference)	Logic diagram, breadboard, block diagram, functional specifications.	Functional specification; logic diagram; circuit diagram; breadboard; test vectors; customer-owned tooling.	Circuit/logic diagram; breadboard; functional specifications; customer-owned tooling.
<b>Design Aids</b>	Calma CAD, logic simulation, circuit simulation.	Logic simulation; breadboard assistance; design rule checks.	Breadboard assistance; cooperative design from logic/circuit diagrams.
<b>Production</b>	In-house, procured	In-house	In-house
<b>Preferred Delivered Product</b>	Packaged units	Probed wafers; packaged dice	Custom packaging; all standard configurations except mapped wafers.
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	Functional, parametric, burn-in, environmental, MIL, linear	Functional; parametric; burn-in	All
<b>Electrical Test Systems Available</b>	Fairchild Sentry	Sentry, Teradyne, Comparator	Fairchild, Datatron, Macrodata, custom.
<b>Comments</b>	MOS: 3 and 5 micron channel lengths.	Tooling can be used by at least two MOS wafer manufacturers to ensure alternate sourcing.	



# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (Cont)

Manufacturer	Micro-Sciences	Mitel Semiconductor (Page 3048)	Monolithic Memories Inc. (Page 4768)
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital	Digital	
Gate Array	CMOS	Silicon-gate CMOS	Field Programmable Logic — PAL and mask — programmed version — HAL.
Chip Density Range (equiv. gates)	300 to 6000	500 — 6000 gates	100 to 350 gates
Cell Library	CMOS, NMOS	Silicon-gate CMOS	
Design Kit Available	No	Yes	Yes
<b>Full Custom Circuits</b>			
Digital	Silicon- or metal-gate PMOS, NMOS, CMOS	Silicon-gate CMOS	
Linear		Silicon-gate CMOS, double poly	
Combined Digital/Linear		Silicon-gate CMOS, double poly	
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Logic diagram; circuit diagram; functional specification; breadboard; customer-owned tooling; known good device.	Functional specification, logic diagram, circuit diagram, customer-owned tooling.	PALS — not required, field programmable. HALs — logic equations, 25-50 "seed" vectors for test generation software.
Design Aids	Logic simulation; breadboard assistance.	Logic simulation, design rule checks.	PALASM software for simulation, fault grading, fuse pattern generation, (mask design input for HALs).
Production	Procured	In-house	In-house
Preferred Delivered Product	Packaged dice	Mapped wafers, probed wafers, scribed dice, packaged dice.	DIPs, LLCs, flat paks
Test Program Generation	Yes	Yes	Yes
Production Test	Functional; parametric; burn-in	Functional, parametric, burn-in, thermal shock, environmental, MIL linear.	Functional, parametric, burn-in, thermal shock, environmental, MIL.
Electrical Test Systems Available		Sentry VII, LTX-DS80	In-house designed
Comments	Chip debugging with portable micro prober on customer's premises.		Several commercial programmers available.

Master Selection Guide

CUSTOM/SEMICUSTOM

**CUSTOM/SEMICUSTOM (Cont)**

<b>Manufacturer</b>	<b>Monosil (Citel) (Page 4628)</b>	<b>Motorola (Page 4796)</b>	<b>Murray Consulting</b>
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	<b>Digital</b>	<b>Digital</b>	<b>Linear, Digital, Combined Linear/Digital</b>
<b>Gate Array</b>	<b>3<math>\mu</math> silicon-gate CMOS CMOS</b>	<b>MCA 600 ECL/1200 ECL; MCA 500 ALS/1300 ALS</b>	<b>CMOS</b>
<b>Chip Density Range (equiv. gates)</b>	<b>200 to 3200</b>	<b>652/1192; 533/1280</b>	<b>1 to 100</b>
<b>Cell Library</b>	<b>Yes</b>	<b>120 functions; 80 functions</b>	<b>BiMOS</b>
<b>Design Kit Available</b>		<b>Manuals</b>	<b>No</b>
<b>Full Custom Circuits</b>			
<b>Digital</b>	<b>Silicon- or metal-gate CMOS, silicon-gate NMOS, bipolar</b>	<b>NMOS, CMOS, ECL</b>	<b>Silicon-gate CMOS; TTL</b>
<b>Linear</b>	<b>Bipolar</b>	<b>No</b>	
<b>Combined Digital/Linear</b>	<b>CMOS</b>	<b>No</b>	<b>Silicon-gate CMOS; bipolar</b>
<b>Provide Design Assistance</b>	<b>Yes</b>	<b>Yes</b>	<b>Yes</b>
<b>Acceptable Customer Input (in order of preference)</b>	<b>Logic diagram; circuit diagram; functional specification; breadboard, data-base tapes, PG tapes, mask sets, customer- owned tooling.</b>	<b>CAD interface, logic simulation, circuit documentation.</b>	<b>Breadboard; circuit diagram; functional specification.</b>
<b>Design Aids</b>	<b>Logic simulation, design rule checks, net-list verification.</b>	<b>Computer aided design package</b>	<b>Design rule checks</b>
<b>Production</b>	<b>In-house</b>	<b>High volume capability</b>	<b>In-house, procured</b>
<b>Preferred Delivered Product</b>	<b>Packaged units; will supply wafers or dice.</b>	<b>Choice of packages</b>	<b>Packaged dice; PC-mounted packaged units.</b>
<b>Test Program Generation</b>	<b>Yes</b>	<b>Yes</b>	<b>No</b>
<b>Production Test</b>	<b>Functional, parametric</b>	<b>100% AC and DC and functional</b>	<b>Functional</b>
<b>Electrical Test Systems Available</b>	<b>Sentry VII</b>	<b>Fairchild Sentry VIII and Series 21</b>	
<b>Comments</b>	<b>Complete custom capability</b>	<b>ECL 10K compatible, ECL 10 KH compatible, TTL-LS compatible, remote design centers.</b>	<b>Develop photo-integrated circuits: CMOS, MOS, bipolar, BiMOS.</b>



# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (Cont)

Manufacturer	NEC Electronics (Page 4799)	NCM Corporation	Nitron
FOR DETAILED DATA SEE:			
Customized Standard Circuits	Digital, Linear, Combined Digital/Linear	Linear, digital, combination	Digital
Gate Array	CMOS, ECL, TTL-LS	CMOS (si-gate, metal-gate) bipolar	Metal-gate CMOS
Chip Density Range (equiv. gates)	To 11,000	1000 (max.)	50 to 600
Cell Library	Yes	YES	Macro library of common functions (overlays)
Design Kit Available	Yes	No	Yes
Full Custom Circuits Digital		CMOS (si-gate, metal-gate) bipolar	Metal-gate PMOS, CMOS
Linear		CMOS (si-gate, metal-gate) bipolar	Metal-gate PMOS, CMOS
Combined Digital/Linear		CMOS (si-gate, metal-gate); bipolar	Metal-gate PMOS, CMOS
Provide Design Assistance		Yes	Yes
Acceptable Customer Input (in order of preference)		Functional specs, logic diagram, circuit diagram, breadboard, test vectors, customer-owned tooling, known good devices.	Customer-owned tooling with debugged test program; logic diagram, electrical specifica- tion with test vectors.
Design Aids		Logic simulation, breadboarding, design rule check, cell library.	All available
Production		Procured	In-house
Preferred Delivered Product		Packaged-units, dice, probed wafers, pc assemblies.	Packaged dice
Test Program Generation		Yes	Yes
Production Test		Functional, parametric, burn-in, thermal cycle.	All commercial and full military
Electrical Test Systems Available		In-house built. $\mu$ c - controlled test systems.	Sentry
Comments			

**CUSTOM/SEMICUSTOM (Cont)**

<b>Manufacturer</b>	<b>National</b>	<b>Pico Design</b>	<b>Plessey Semiconductors</b> <b>(Page 4815)</b>
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital	No	Digital, Linear, Combined Digital/Linear
Gate Array	Silicon-gate CMOS	No	Bipolar, ECL and CMOS
Chip Density Range (equiv. gates)	To 2400	10,000	ECL to 1200, CMOS to 10440
Cell Library	Yes	Yes	Yes
Design Kit Available	Yes	No	Yes
<b>Full Custom Circuits</b>			
Digital	Metal-gate PMOS, metal-or Silicon-gate NMOS, high voltage metal-gate CMOS, low voltage metal-and silicon-gate CMOS, metal double poly silicon-gate CMOS, dual layer XMOS, oxide isolated and Schottky devices, I <sup>2</sup> L.	Silicon- or metal-gate PMOS, NMOS, CMOS, other MOS.	PMOS, NMOS, CMOS; I <sup>2</sup> L, ECL
Linear	Same as above	No	Bipolar, CMOS
Combined Digital/Linear	Same as above	No	I <sup>2</sup> L, ECL Bipolar, CMOS
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Any	Logic diagram, circuit schematic, breadboard, functional specification.	Logic diagram, breadboard, customer-owned tooling, data base tape.
<b>Design Aids</b>	Computer aided circuit analysis, logic and systems simulation, digitizing cell plotting and editing, design rule check, pattern generation, photolithography.	Logic and circuit simulation; design rule checks; CALMA GDS II.	Design rule check, testability analyzer, fault analyzer, auto-placement, auto-router, logic simulator.
<b>Production</b>	In-house	Procured	
<b>Preferred Delivered Product</b>	Packaged dice	All except PC-mounted packaged units.	Packaged tested dice, scribed dice, probed wafers, mapped wafers.
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	All except linear	Functional; parametric	Functional, parametric, burn-in, MIL-STD-883.
<b>Electrical Test Systems Available</b>	Century VI and VII, Sentinel, Teradyne, Megatest.	Fairchild, Sentry VII and Series 20	Teradyne J274, Fairchild, Sentry special in-house.
<b>Comments</b>			

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (Cont)

Manufacturer	Polycore Electronics Inc.	Precision Monolithics, Inc.	Raytheon (Page 4822)
FOR DETAILED DATA SEE:			
Customized Standard Circuits		Linear	Linear, Digital
Gate Array		From customer-owned tooling	Bipolar digital ISL
Chip Density Range (equiv. gates)			800 to 2400
Cell Library		Standard parts available in chip form	Yes, TTL
Design Kit Available		Kit parts	Yes
Full Custom Circuits Digital	CMOS metal-gate	TTL, LSTTL, ISL, I <sup>2</sup> L, ECL, ISO/CMOS-SiGate	Consult factory (bipolar)
Linear	Bipolar	20, 40, 60V supply voltage	Bipolar
Combined Digital/Linear	Power Interface/Driver Circuits	I <sup>2</sup> L, ECL, TTL, LSTTL	Consult factory (bipolar)
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Schematics, Performance specs.	Customer-owned tooling (pattern-generator-tape, composite drawing), known good device.	TEGAS net list, circuit schematics, data base program.
Design Aids	Breadboard simulation	Kit parts	Evaluation devices, design source manual, factory tutorial.
Production	In-house wafer fabrication procurred assembly	In-house 3", 4" lines	Yes
Preferred Delivered Product	Packaged units, dice	Mapped wafers, probed wafers, packaged units.	
Test Program Generation	Yes	Yes	Yes
Production Test	Functional, parametric, burn-in; thermal shock, environmental, MIL.	Functional, parametric, burn-in, thermal shock, environmental, MIL, linear.	Yes
Electrical Test Systems Available	MCT 200	LTX, F-5000	Sentry 7, Sentry Series 21
Comments	Also provide silicon foundry service in linear, I <sup>2</sup> L and CMOS metal gate.	Ion implantation; dual layer metallization; nitride passivation; thin film resistors; MIL 35810 qualified.	

**CUSTOM/SEMICUSTOM (cont)**

<b>Manufacturer</b>	<b>RCA</b>	<b>Semi Processes, Inc.</b>	<b>Si Fab</b>
<b>FOR DETAILED DATA SEE:</b>	<b>(Page 4824)</b>	<b>(Page 4826)</b>	<b>(Page 4827)</b>
<b>Customized Standard Circuits</b>	Digital	Digital	Digital
Gate Array	Silicon-gate, metal-gate CMOS	Silicon-gate CMOS	Silicon- and metal-gate
Chip Density Range (equiv. gates)	168 to 840	48 to 1400	
Cell Library	CMOS	Yes	Yes
Design Kit Available	No	Yes	
<b>Full Custom Circuits</b>			
Digital	Silicon-gate, metal-gate CMOS		CMOS
Linear			CMOS
Combined Digital/Linear			CMOS
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Logic diagram, circuit diagram, breadboard test vectors, customer-owned tooling (pattern-generator tape, composite drawing), known good device.	Logic diagram; circuit diagram; test vectors; custom-owned tooling; breadboard; functional specification.	
Design Aids	Logic simulation; breadboard assistance; design rule checks.	Design rule checks, logic simulation, circuit simulation	
Production	In-house	In-house	In-house
Preferred Delivered Product	Probed wafers, scribed dice, packaged dice.	As requested	
Test Program Generation	Yes	Yes	
Production Test	Functional, parametric, burn-in	Functional; parametric; burn-in, ac	
Electrical Test Systems Available	Teradyne; Century 7; J273, J193, J283, J325; LTX; Tester; MTS 77.	Sentry, Genrad, Pragmatic	
Comments		Breadboard assistance using SPI's proprietary 74HC low power CMOS logic IC's.	



# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Signetics (Page 4829)	Signetics (Page 4829)	Silicon Systems (Page 4872)
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital	Digital	
Gate Array	Silicon-gate CMOS	ECL 100K and 10 K compatibility (0.5 ns)	
Chip Density Range (equiv. gates)	330 to 1100 1250 to 5000 (mid '84)	500-800	
Cell Library	Silicon-gate CMOS, TTL and LSTTL	Macrocell Library	
Design Kit Available	Yes	Yes	
<b>Full Custom Circuits</b>			
Digital	No*		Silicon- or metal-gate PMOS, NMOS, CMOS, bipolar, TTL, STL, SRTL, LSTTL, ECL, I <sup>2</sup> L
Linear	No		Silicon- or metal-gate PMOS, NMOS, CMOS, bipolar
Combined Digital/Linear	TBD		Silicon- or metal-gate CMOS, bipolar, TTL, SRTL, I <sup>2</sup> L, STL
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Logic diagram, timing diagram, and simulation output; CAD output.	Calma tapes plus simulation results; Mylar symbolic representation plus simulation results.	Functional specification, logic diagram, circuit diagram, breadboard, test vectors.
Design Aids	Schematic capture, complete simulation, auto place, auto route, auto test generation.	Logic simulation, CAD design rule checks, layout to simulation check, test generation, design manual, training sessions.	Breadboards, logic and circuit simulation, design-rule checks, test-program development.
Production	In-house	In-house	Procured and in-house
Preferred Delivered Product	Packaged units	Packaged units	Packaged units and tested dice
Test Program Generation	Yes	Yes	Yes
Production Test	Functional, parametric, burn-in	Functional, parametric and ac	Automatic testers for analog and digital devices.
Electrical Test Systems Available	Sentry VII, VIII, and 21	Tektronix (25 MHz)	
Comments	* Standard cells from 200 to 2000 gates.		

**CUSTOM/SEMICUSTOM (cont)**

Manufacturer	Siltronics	Solid State Scientific	Sprague Electric Co.
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>			Linear
Gate Array			
Chip Density Range (equiv. gates)			
Cell Library			
Design Kit Available			No
<b>Full Custom Circuits</b> Digital	TTL, STTL, LSTTL, I <sup>2</sup> L, ECL, other bipolar	Silicon-gate CMOS, NMOS metal-gate CMOS	
Linear	TTL, STTL, LSTTL, I <sup>2</sup> L, ECL, other bipolar	Silicon-gate CMOS, NMOS	
Combined Digital/Linear	TTL, STTL, LSTTL, I <sup>2</sup> L, ECL, other bipolar	Silicon-gate NMOS, CMOS	
Provide Design Assistance		Yes	
Acceptable Customer Input (in order of preference)		Minimum of functional diagram; breadboard; pattern generator tape or database tape.	Composite drawing
<b>Design Aids</b>	Breadboard assistance; design-rule checks.	Logic simulator assistance, breadboard assistance, design rule checks.	
<b>Production</b>	Breadboard wafers; in-house assembly.	In-house	In-house
<b>Preferred Delivered Product</b>	Fully tested assembled package.	No preference	Wafers, dice, packaged devices
<b>Test Program Generation</b>	Yes	Yes	
<b>Production Test</b>	Customized to suit	Full screening available including burn-in and full environmental screening.	
<b>Electrical Test Systems Available</b>	J259-style equipment; custom testers.	Sentry VII, Sentinal, Teradyne	
<b>Comments</b>	Specialize in mixed analog/ digital bipolar. Die sizes to 40,000 mil <sup>2</sup> .	Capabilities include: design, layout, CAP, mask shop, wafer fabrication assembly, and test.	ULN-2350C and ULN-2351C basic unmetallized circuit.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Standard Microsystems	Sunshine Semiconductor	Supertex
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>	Digital, Combined Digital/Linear	Digital, Combined Linear/Digital	Linear, Digital
Gate Array			Silicon- or metal-gate CMOS
Chip Density Range (equiv. gates)			50 to 2000
Cell Library	Silicon-gate NMOS, CMOS	Si-gate CMOS	
Design Kit Available	No		No
<b>Full Custom Circuits</b>			
Digital	Metal-gate PMOS; silicon-gate NMOS, CMOS	Silicon-gate CMOS, metal-gate CMOS, nonvolatile CMOS	Silicon- or metal-gate CMOS
Linear			Silicon- or metal-gate CMOS
Combined Digital/Linear	Silicon-gate NMOS, CMOS	Metal Gate CMOS	Silicon- or metal-gate CMOS
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Functional specification, logic diagram, Calma database tape, other.	Product description, block diagram with specifications, logic diagram.	Breadboard with functional specification and logic diagram; customer-owned tooling.
<b>Design Aids</b>	Logic simulation; breadboard design rule checks, electrical rule checks, transient analysis	Breadboard, circuit and logic simulation, DRC, graphics, NCC	Logic simulation; breadboard assistance.
<b>Production</b>	In-house	Procured	In-house
<b>Preferred Delivered Product</b>	Packaged devices preferred. All others available.	Packaged dice, bare dice, design and layout	Probed wafers; inspected dice; packaged dice.
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	Full testing and screening, including burn-in.	Procured	Functional; parametric; burn-in; environmental.
<b>Electrical Test Systems Available</b>	Megatest Q8000; SMC; dedicated testers, Fairchild, Sentry and Sentinal.		Teradyne J193; GenRad 2225
<b>Comments</b>	Specialize in digital MOS/LSI and VLSI; will also customize by modifying MOS/LSI standard parts. Second sources available.	Services include product and specification finalization, logic and circuit design, layout, test definition.	

**CUSTOM/SEMICUSTOM (cont)**

Manufacturer	Synertek	Telmos	Telmos
FOR DETAILED DATA SEE:	(Page 4873)	(Page 4877)	(Page 4877)
Customized Standard Circuits		Digital	Digital/Linear
Gate Array		Silicon-gate and metal-gate CMOS	Silicon-gate analog/digital Silicon-gate high voltage
Chip Density Range (equiv. gates)		50 to 1260 gates	
Cell Library		Silicon-gate and metal-gate CMOS digital	100 to 1200 gates
Design Kit Available	Yes		Yes
Full Custom Circuits			
Digital	Silicon-gate CMOS, NMOS	Custom designs are preferred	
Linear	Same as above		
Combined Digital/Linear	Same as above	Silicon-gate analog/digital	Silicon-gate analog/digital
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Functional specification, customer owned tooling, logic diagram, circuit diagram, breadboard.	Schematic I/O requirements, functional spec, breadboard, test vectors, customer owned tooling, known good device.	Schematic I/O requirements, functional specs, breadboard test vectors, customer owned tooling, known good device.
Design Aids	Logic simulation breadboard assistance, continuity check, design rule check.	Aspec, Spice IIG — transient analysis logic, Tegas IV — logic simulation, Sentry VII — test.	Aspec, Spice IIG, transient analysis logic, Tegas IV, logic simulation.
Production	In-house	In-house wafer fab facility	In-house wafer fabrication
Preferred Delivered Product	Packaged dice, scribed dice	Wafer, dual in line packages, chip carriers.	Wafer, dual-in-line packages, chip carriers.
Test Program Generation	Yes	Yes	Yes
Production Test	All	Functional, parametric burn-in thermal shock, environmental.	Functional, parametric burn-in thermal shock, environmental
Electrical Test Systems Available	Century, Xincom		Sentry VII
Comments	Customer interface ranges from "black box" specification to COT. Company will work with customer engineers to develop in-house capability for designing future COT.		



# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	Texas Instruments	TLSI, Inc.	Torric Corporation
FOR DETAILED DATA SEE:	(Page 4878)		
Customized Standard Circuits	Digital		Linear, Digital, Combined Digital/Linear
Gate Array	LPS, silicon-gate CMOS, STL	Silicon-gate CMOS	Silicon or metal-gate CMOS, Linear Bipolar, Digital ECL
Chip Density Range (equiv. gates)	540 to 4000	300 + 2500	300 — 2500
Cell Library	Software functions — bipolar CMOS MACROS	Si-gate CMOS, NMOS	Silicon-gate CMOS
Design Kit Available	Yes	No	Yes
Full Custom Circuits			
Digital	LPS, STL, PL, CMOS, NMOS	CMOS, NMOS, PMOS	Silicon- or metal-gate PMOS, NMOS, CMOS, TTL, STTL, PL
Linear	Yes	CMOS, NMOS, PMOS	Silicon- or metal-gate CMOS; TTL, STTL
Combined Digital/Linear	Yes	CMOS, NMOS, PMOS	Silicon- or metal-gate CMOS; TTL, STTL
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Logic diagram, PG tape, HDL/TDL design input, functional specification.	Specification, logic diagram schematic, breadboard COT	Functional specification; logic diagram; circuit diagram; breadboard; known good device; customer owned tooling; test vectors.
Design Aids	Simulation, testability analysis, test grading, load checker. Texas Instruments logic array design utility — TILADS.	Custom cell library, logic simulation, CAD, CAE, Design Rule checks.	Logic simulation; breadboard assistance; design rule checks; graphics CAD.
Production	In-house	Procured	Procured
Preferred Delivered Product	Packaged devices	Package die, dice	Packaged dice, printed-card-mounted packaged units.
Test Program Generation	Yes	Yes	Yes
Production Test	Functional, parametric, burn-in, MIL.	Functional, parametric, burn-in, linear	No
Electrical Test Systems Available	Sentry 20, in-house	LTX, HP, Gen Rad, dedicated Multiple production sources;	
Comments		no minimum production requirements.	

**CUSTOM/SEMICUSTOM (cont)**

<b>Manufacturer</b>	<b>Trans-Data</b>	<b>Universal Semiconductor (Page 4930)</b>	<b>VLSI Design Associates</b>
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>		Digital	Digital/Analog
Gate Array		Silicon-gate ISO-CMOS	
Chip Density Range (equiv. gates)		180 to 2400	
Cell Library		Silicon-gate ISO-CMOS	Silicon-gate CMOS, Digital and Analog
Design Kit Available		Yes	No
<b>Full Custom Circuits Digital</b>	Silicon-gate NMOS and CMOS	Silicon-gate ISO-CMOS	Silicon-gate CMOS and NMOS; logic and memory; nonvolatile memory.
Linear	Silicon-gate — CMOS	Same as above	Silicon-gate CMOS and NMOS; CCD; Switched capacitor filters.
Combined Digital/Linear	Silicon-gate — CMOS	Same as above	
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Functional specs, logic diagram, circuit diagram, known good device, customer tooling.	Logic reconfigured at GATE/FF Level. Logic diagram, PG tapes, masks.	Logic diagram, circuit diagram, functional specification, breadboard.
<b>Design Aids</b>		Logic simulation, breadboard assistance, design rule checks. Auto test generation; auto place/route; logic to layout verification.	In-house: Calma GDS II and VAX11/780.
<b>Production</b>	Procured	In-house	Procured
<b>Preferred Delivered Product</b>	Packaged dice (prototype quantities), probed wafers.	Package parts, dice probed wafers.	As requested
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>		Functional, parametric, burn-in, thermal shock, environmental, MIL.	Functional, parametric
<b>Electrical Test Systems Available</b>	Functional, Parametric, Burn-in, Thermal Shock, environmental MIL.	Micro Manipulator, Sentry II, VI, VII, Sentry System 20, pragmatic functional tester. Fairchild Series 10.	
<b>Comments</b>	Provide assistance to customers who will utilize gate arrays manufactured by others.	With smaller random logic circuits, prefer to start with a gate array and convert to full custom when manufacturing volumes justify the conversion.	Independent design center; multiple sources, design aids.

# MASTER SELECTION GUIDE

## CUSTOM/SEMICUSTOM (cont)

Manufacturer	VLSI Technology (Page 4931)	Western Digital	ZyMOS
<b>FOR DETAILED DATA SEE:</b>			
<b>Customized Standard Circuits</b>		Digital	Digital, combined digital/linear
Gate Array		Up to 1000 gates, Silicon-gate NMOS	Up to 2500 gates, metal-gate CMOS 3000 gates, silicon-gate CMOS 4000 gates, NMOS.
Chip Density Range (equiv. gates)			Metal-gate CMOS, silicon-gate CMOS, NMOS
Cell Library		Yes	Yes
Design Kit Available	NMOS, HMOS, CMOS		Metal-gate CMOS, silicon-gate CMOS, NMOS
<b>Full Custom Circuits</b>			
Digital			
Linear	NMOS, CMOS		CMOS
Combined Digital/Linear	NMOS, CMOS		CMOS
Provide Design Assistance	Yes	Yes	Yes
Acceptable Customer Input (in order of preference)	Functional specification logic diagram, breadboard, cus- tomer-owned tooling, data base tape, PG tape, mask.	Functional specification logic system	Logic input through ZyP CAD system, logic diagram, functional specification, customer-owned tooling.
<b>Design Aids</b>	PRISM VLSI design system, in- cluding high-level design lan- guage, circuit extraction logic and circuit simulation, design rule checks, plotting, STIX.	Logic simulation, design rule checks, test program development.	Circuit simulation, logic simulation, test program generation, artwork generation all linked to common data base, design rule checks, breadboard assistance.
<b>Production</b>	In-house	In-house	In-house
<b>Preferred Delivered Product</b>	Packaged dice preferred. Any on request.	Packaged units and tested dice	Packaged dice preferred, any on request
<b>Test Program Generation</b>	Yes	Yes	Yes
<b>Production Test</b>	Functional, parametric, burn-in, thermal shock, environmental, MIL.	Functional, burn-in	Functional, parametric
<b>Electrical Test Systems Available</b>	Sentry 20/120, Accutest		Sentry, VII, Lomac
<b>Comments</b>	Multiple sourcing compatibility, foundry services, training and CAD aids to support User- Designed VLSI and Mead- Designed VLSI.	Uncommitted logic arrays, 20 pin DIP, 130 prefabricated logic elements or 28/40 pin DIP, 400 prefabricated logic elements.	ZyP CAD system — logic level design at customer's facility, CMOS and NMOS silicon level simulation.



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CUSTOM/SEMICUSTOM-Gate Arrays

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line				
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL									
<b>Gate Arrays</b>																			
—	15		1200			80	x	x	x		4 to 10		H1084	† Hughes					
0.5	300	ECL	2800		(68)	120					x	-4.5	Macrocell array	<b>MCA2500ECL</b>	<b>Motorola (4796)</b>	5			
					(68)	120							x	-4.5	Macrocell array		MCA2500ECL	National	
		ECL-ALS	1456	Dir.	49	99		x	x	x		-4.5	1280-bit RAM on-chip	MCA1500M	National				
450	ECL		638	30		28		x	x	x		-4.5 to -5.2		<b>ACE600</b>	<b>Signetics (4858)</b>				
			878	42		32			x	x	x		-4.5 to -5.2		<b>ACE900</b>	<b>Signetics (4858)</b>			
			1000			96			x	x	x		-4.5 to -5.2	320-bit RAM on-chip	<b>ACE1320M</b>	<b>Signetics (4860)</b>			
			1414			96			x	x	x		-4.5 to -5.2		<b>ACE1400</b>	<b>Signetics (4860)</b>			
			2204			128			x	x	x		-4.5 to -5.2		<b>ACE2200</b>	<b>Signetics (4860)</b>	10		
500	ECL		20			21				x		-4.5 to -5.7	20 equiv. 3-input gates	<b>FGE0020</b>	<b>Fairchild (4634)</b>				
1000	ECL		500			72		x	x	x		-4.2 to -5.7 or 5	500 equiv. 3-input gates	<b>FGE0500</b>	<b>Fairchild (4634)</b>				
			1000			86		x	x	x		-4.2 to -5.7 or 5	1000 equiv. 3-input gates	FGE1000	Fairchild				
			2000			126		x	x	x		-4.2 to -5.7 or 5	2000 equiv. 3-input gates	<b>FGE2000</b>	<b>Fairchild (4634)</b>				
0.55	400	ECL	75			24					x	-5.2		SCD1000	Plessey	15			
				300	36		20					x	-5.2		SCD2000H		Plessey		
				600			54						x	-5.2			<b>SCD4000</b>	<b>Plessey (4815)</b>	
0.75	700	ECL	300	28		28						x	-4.5		<b>μPB6301</b>	<b>NEC (4799)</b>			
			500	1200	40		48							x	-4.5		<b>μPB6310</b>	<b>NEC (4799)</b>	
1.05	500	ECL	2000	60		48							x	-4.5		<b>μPB6320</b>	<b>NEC (4799)</b>	20	
1.2	85	TTL-AS	1500			88		x	x					5, 2		HGA-B01500	Harris (4741)		
			3000			120			x	x					5, 2		HGA-B02500	Harris (4741)	
160	ECL		625	(18)		60						x	-5.2	Macrocell array	MCA600	AMD	25		
				(18)		60							x	-5.2	Macrocell array	LCA600		LSILogic (4762)	
				(18)		60							x	-5.2	Macrocell array	<b>MCA600ECL</b>		<b>Motorola (4796)</b>	
				(18)		60							x	-5.2	Macrocell array	MCA600ECL		National	
			1192	(26)		60								x	-5.2	Macrocell array		MCA1200	AMD
				(26)		60								x	-5.2	Macrocell array		LCA1200	LSILogic (4762)
				(26)		60								x	-5.2	Macrocell array		<b>MCA1200ECL</b>	<b>Motorola (4796)</b>
				(26)		60								x	-5.2	Macrocell array		MCA1200ECL	National
200	ECL		250	8		26					x	x	-5.2/5	Mixed TTL/ECL outputs	Q720	† AppMicroCir	30		
			500	28		28						x	x	-5.2/5	Mixed TTL/ECL outputs	Q710		† AppMicroCir	
			1000	38		38						x	x	-5.2/5	Mixed TTL/ECL outputs	Q700		† AppMicroCir	
			1500	46		38						x	x	-5.2/5	Mixed TTL/ECL outputs	Q750		† AppMicroCir	
1.4	160	ECL	2800			120		x	x				5	Macrocell array	<b>MCA2800ALS</b>	<b>Motorola (4797)</b>	35		
						120		x	x						5	Macrocell array		MCA2800ALS	National
1.5	75	TTL-AS	1500			88		x	x				5, 2		HGABA01500	† Harris	40		
			3000			120			x	x				5, 2		HGABA03000		† Harris	
200	ECL		250	8		26					x	x	-5.2/5	Mixed TTL/ECL outputs	Q720	† AppMicroCir	40		
			500	28		28						x	x	-5.2/5	Mixed TTL/ECL outputs	Q710		† AppMicroCir	
			1000	38		38						x	x	-5.2/5	Mixed TTL/ECL outputs	Q700		† AppMicroCir	
			1500	46		38						x	x	-5.2/5	Mixed TTL/ECL outputs	Q750		† AppMicroCir	
250	ECL		300	36		20						x	-5.2		SCD2000M	Plessey			
1.5 *	200	TTL-LS	2000			112		x	x			5, 2, 3	2108 equiv. 3-input gates	<b>B2000</b>	<b>Fajitsu (4638)</b>	45			
1.6	50	CMOS	1500			76		x	x	x			3 to 10		BC210	† Barvon	45		
			2500			108		x	x	x				3 to 10		BC225		† Barvon	
1.8	40	CMOS	968			54		x	x	x			3 to 6		<b>LL7090</b>	<b>† LSILogic (4762)</b>	50		
			1430			68		x	x	x				3 to 6		<b>LL7140</b>		<b>† LSILogic (4762)</b>	
			2224			86		x	x	x				3 to 6		<b>LL7220</b>		<b>† LSILogic (4762)</b>	
			3268			110		x	x	x				3 to 6		<b>LL7320</b>		<b>† LSILogic (4762)</b>	
			4368			122		x	x	x				3 to 6		<b>LL7430</b>		<b>† LSILogic (4762)</b>	
			6425			160		x	x	x				3 to 6		<b>LL7640</b>		<b>† LSILogic (4762)</b>	

† Military Temperature Range (-55° to 125°C)

\* Typical Value  
**Bold face indicates additional data is provided on the page noted.**

(Continued)

Master Selection Guide  
CUSTOM/SEMICUSTOM

CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line	
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL						
<b>Gate Arrays</b>															<b>(Cont'd)</b>	
1.8	40	CMOS	8456			180	x	x	x		3 to 6		LL7840	†LSILogic	(4762)	5
			10013			180	x	x	x		3 to 6		LL71000	†LSILogic	(4762)	
	45	CMOS	500			40	x	x	x		2 to 9		IGD10500	†Intersil		
			1000			68	x	x	x		2 to 9		IGD11000	†Intersil		
			1500			84	x	x	x		2 to 9		IGD11500	†Intersil		
			2600			100	x	x	x		2 to 9		IGD12600	†Intersil		
			4500			140	x	x	x		2 to 9		IGD14500	†Intersil		
6000			180	x	x	x		2 to 9		IGD16000	†Intersil					
1.9 *	—	TTL-LS	360			48			x	x	5	360 equiv. 3-input gates	B350	Fujitsu	(4637)	10
			1120			88				x	x	5	1120 equiv. 3-input gates	B1100	Fujitsu	
2.0	50	CMOS	2500	39		84	x	x			5		F6C2400	†Fairchild	(4634)	15
			3925	31		90	x	x			5		F6C4000	†Fairchild	(4634)	
			6006	41		120	x	x			5		F6C6000	†Fairchild	(4634)	
2.1	110	CMOS	1260	17		42	x	x	x		4 to 6		SCX6212	National		
			2385	55		56	x	x	x		4 to 6		SCX6224	National		
			2430	12		76	x	x	x		4 to 6		SCX6225	National		
			4860	53		54	x	x	x		4 to 6		SCX6248	National		
			6260	66		88	x	x	x		4 to 6		SCX6260	National		
2.2	160	TTL-AS	1280	(88)	(44)	88				x	5, 2	Mixed TTL/ECL compatible	TAT010	†TI		
				(88)	(44)	88					x	-5.2		TAT010	†TI	
			2420	(120)	(60)	120				x		5, 2	Mixed TTL/ECL compatible	TAT020	†TI	
				(120)	(60)	120					x	-5.2		TAT020	†TI	
2.2 *	—	TTL-LS	500			62			x	x	5	512 equiv. 3-input gates	B500	Fujitsu	(4637)	
2.4	50	CMOS	360			36	x	x	x		3 to 6		IS03A	†Siliconix		
						36	x	x	x		3 to 6		IS03A	†Universal	(4930)	
			540			48	x	x	x		3 to 6		IS03B	†Siliconix		
						48	x	x	x		3 to 6		IS03B	†Universal	(4930)	
			720			54	x	x	x		3 to 6		IS03C	†Siliconix		
						54	x	x	x		3 to 6		IS03C	†Universal	(4930)	
			960			62	x	x	x		3 to 6		IS03D	†Siliconix		
						62	x	x	x		3 to 6		IS03D	†Universal	(4930)	
			1200			68	x	x	x		3 to 6		IS03E	†Siliconix		
						68	x	x	x		3 to 6		IS03E	†Universal	(4930)	
			1500			70	x	x	x		3 to 6		IS03F	†Siliconix		
						70	x	x	x		3 to 6		IS03F	†Universal	(4930)	
			1800			82	x	x	x		3 to 6		IS03G	†Siliconix		
						82	x	x	x		3 to 6		IS03G	†Universal	(4930)	
2400			82	x	x	x		3 to 6		IS03H	†Siliconix					
			82	x	x	x		3 to 6		IS03H	†Universal	(4930)				
2.5	200	ECL	300	36		20					x	-5.2	SCD2000L	†Plessey		
3.0	—	STL	560			61			x	x			STL700	†TI	(538)	
						28	x	x	x		4 to 6		MA0200	†Citel	(4628)	
	25	CMOS	228			28	x	x	x		4 to 6		MA0200	Matra		
						36	x	x	x		4 to 6		MA0400	†Citel	(4628)	
			380			36	x	x	x		4 to 6		MA0400	Matra		
						50	x	x	x		4 to 6		MA0800	†Citel	(4628)	
			754			50	x	x	x		4 to 6		MA0800	Matra		
						52	x	x	x		2.5 to 5.5		GA1000	†AMI	(4619)	
	1139			62	x	x	x		4 to 6		MA1200	†Citel	(4628)			
	1260	1		62	x	x	x		4 to 6		MA1200	Matra				
	1260	1		75	x	x	x		3 to 7		HCS12000	†CalDevices				
	1500			64	x	x	x		2.5 to 5.5		GA1500	†AMI	(4619)			
	1782	1		89	x	x	x		3 to 7		HCS17800	†CalDevices				
	2025			74	x	x	x		2.5 to 5.5		GA2000	†AMI	(4619)			
2508			84	x	x	x		2.5 to 5.5		GA2500	†AMI	(4619)				

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line	
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL						
<b>Gate Arrays</b>															<b>(Cont'd)</b>	
3.0															(Cont'd)	
30	CMOS		1000			62	x	x	x		2.5 to 5.5		GA1000D	†AMI (4619,4619)	5	
			2070			84	x	x	x		2.5 to 5.5		GA2000D	†AMI (4619,4619)		
			3080			102	x	x	x		2.5 to 5.5		GA3000D	†AMI (4619)		
			4012			120	x	x	x		2.5 to 5.5		GA4000D	†AMI (4619)		
			4995			134	x	x	x		2.5 to 5.5		GA5000D	†AMI (4619)		
35	CMOS		400			40	x	x	x		3 to 8		SGA04	† MicroTech	10	
			600			48	x	x	x		3 to 8		SGA06	† MicroTech		
			1000			58	x	x	x		3 to 8		SGA10	† MicroTech		
			1500			72	x	x	x		3 to 8		SGA15	† MicroTech		
			2000			82	x	x	x		3 to 8		SGA20	† MicroTech		
			3000			110	x	x	x		3 to 8		SGA30	† MicroTech		
			4000			120	x	x	x		3 to 8		SGA40	† MicroTech		
			5000			138	x	x	x		3 to 8		SGA50	† MicroTech		
			6000			168	x	x	x		3 to 8		SGA60	† MicroTech		
50	ECL		533	26	27			x	x		5	Macrocell array	MCA500ALS	Motorola (4797)		15
				26	27			x	x		5	Macrocell array	MCA400ALS	National		
			1280	40	40			x	x		5	Macrocell array	MCA1300ALS	Motorola (4797)		
				40	40			x	x		5	Macrocell array	MCA1300ALS	National		
85	CMOS		648			56	x	x	x		5		HGA-C06600	† Harris (4740)	20	
			1296			80	x	x	x		5		HGA-C01200	† Harris (4740)		
			2520			102	x	x	x		5		HGA-C02500	† Harris (4740)		
100	CMOS		6000			120	x	x	x		3 to 7		CLA4000	Plessey	25	
125	TTL		1000			76		x	x		5,2,5		FST0750	Fairchild (4634)		
130	ECL		300	36		20				x	-5.2		SCD2000L	† Plessey		
3.2	55	ISL	836			48		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA8L48	† Raytheon (508,4822)		
			1196			60		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA12L60	† Raytheon (508,4822)		
			1620			68		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA16L68	† Raytheon (508,4822)		
			1984			76		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA20L76	† Raytheon (508,4822)		
			2376			84		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA24L84	† Raytheon (508,4822)		
3.5	20	CMOS	500			44	x	x	x		4 to 6		50500B	GTEMicro	30	
			1000			54	x	x	x		4 to 6		51000B	GTEMicro		
			1500			68	x	x	x		4 to 6		51500B	GTEMicro		
			2000			78	x	x	x		4 to 6		52000B	GTEMicro		
50	CMOS		150			32	x	x	x		3 to 7		AT-HSD-150	† ArrayTech	35	
			250	18	12	32	x	x	x		3 to 7		AT-HSD-250	† ArrayTech		
			350			36	x	x	x		3 to 7		AT-HSD-350	† ArrayTech		
			550			48	x	x	x		3 to 7		AT-HSD-550	† ArrayTech		
			750			54	x	x	x		3 to 7		AT-HSD-750	† ArrayTech		
			1000			62	x	x	x		3 to 7		AT-HSD-1000	† ArrayTech		
			1200			68	x	x	x		3 to 7		AT-HSD-1200	† ArrayTech		
			1500			76	x	x	x		3 to 7		AT-HSD-1500	† ArrayTech		
			1800			82	x	x	x		3 to 7		AT-HSD-1800	† ArrayTech		
		ISL	836			48		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA8L48	† Raytheon		
			1196			60		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA12L60	† Raytheon		
			1620			68		x	x		5,1,8	Gate delay 2.3 ns max with active pull-up	CGA16L68	† Raytheon		

† Military Temperature Range (-55° to 125°C)

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CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line	
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL						
<b>Gate Arrays</b>															<b>(Cont'd)</b>	
3.5	50	ISL	1984			76			x	x		5, 1.8	Gate delay 2.3 ns max with active pull-up	CGA20L76	† Raytheon	(Cont'd)
			2376			84			x	x		5, 1.8	Gate delay 2.3 ns max with active pull-up	CGA24L84	† Raytheon	
	66	CMOS	648			37	x	x	x			4 to 6		SCX6306	National	
			1260	17		42	x	x	x			4 to 6		SCX6312	National	
			2385	55		56	x	x	x			4 to 6		SCX6324	National	
			2430	12		76	x	x	x			4 to 6		SCX6325	National	
			4860	54		54	x	x				4 to 6		SCX6348	National	
			6090	66		88	x	x				4 to 6		SCX6360	National	
	75	CMOS	648			56	x	x	x			5		HGA-C00600	† Harris (4740)	
			1296			80	x	x	x			5		HGA-C01200	† Harris (4740)	
			2520			102	x	x	x			5		HGA-C02500	† Harris (4740)	
	80	STL	540	(76)	(38)	76			x			5, 2		TAT004	† TI (538.4894)	
			1008	(104)	(52)	104			x			5, 2		TAT008	† TI (538.4894)	
3.5 *	25	CMOS	300	1		37	x		x	x		3 to 7		HCS3100	† CalDevices	
			400	1		43	x		x	x		3 to 7		HCS4100	† CalDevices	
			500			46	x	x	x			2.5 to 5.5		GA500	† AMI (4619)	
			540	1		49	x	x	x			3 to 7		HCS5400	† CalDevices	
			770	1		59	x	x	x			3 to 7		HCS7700	† CalDevices	
			1000	1		67	x	x	x			3 to 7		HCS10000	† CalDevices	
3.6	100	TTL-LS	1000	66		52			x	x		5		μPB6110	NEC	
3.8	75	CMOS	8000			160			x	x		5		C8000VH	Fujitsu (4640)	
4.0	12	CMOS	330			38	x	x	x			2 to 6		SCC0335H	Signetics (4866)	
			448			26	x	x	x			2 to 6		SCC0455H	Signetics (4866)	
			704			38	x	x	x			3 to 15		SCC0705H	Signetics (4866)	
			1106			66	x	x	x			2 to 6		SCC1105H	Signetics (4866)	
	15	CMOS	500			50	x	x	x			5		HD61J	Hitachi	
			1000			68	x	x	x			5		HD61K	Hitachi	
			1600			68	x	x	x			5		HD61L	Hitachi	
			2500			104	x	x	x			5	Optional 256-bit RAM	HIT	Hitachi	
	20	CMOS	504			38	x	x	x			0.5 to 7		Interchip 50	† IntMicCir	
			960			48	x	x	x			0.5 to 7		Interchip 100	† IntMicCir	
			1444			62	x	x	x			0.5 to 7		Interchip 150	† IntMicCir	
			2000	(76)	(76)		x	x	x			3 to 6.5		MA4200	† Mitel	
	25	ISL	1192			36			x	x		5, 1.5		8A1200	† Signetics (4834)	
						60			x	x		5, 1.5		8A1260	† Signetics (527.4844)	
			1472			42			x	x		5, 1.5		8A1542	† Signetics (527.4848)	
			1620			64			x	x		5, 1.5		8A1664	† Signetics (527.4852)	
			1740			64			x	x		5, 1.5		8A1864	† Signetics (527.4853)	
			2088			76			x	x		5, 1.5		8A2176	† Signetics (527.4854)	
	35	CMOS	4104			118	x	x	x			5		μPD65040	NEC	
			6528			134	x	x	x			5		μPD65060	NEC	
			11250			172	x	x	x			5		μPD65100	NEC	
	38	CMOS	135			32	x	x	x			3 to 10		G70090	† IntMicCir (4747)	
			300			40	x	x	x			3 to 10		G70200	† IntMicCir (4747)	
			375			40	x	x	x			3 to 10		G70250	† IntMicCir (4747)	
			540			52	x	x	x			3 to 10		G70360	† IntMicCir (4747)	
			735			58	x	x	x			3 to 10		G70490	† IntMicCir (4747)	
			960			68	x	x	x			3 to 10		G70640	† IntMicCir (4747)	
			1215			76	x	x	x			3 to 10		G70810	† IntMicCir (4747)	
			1500			84	x	x	x			3 to 10		G71000	† IntMicCir (4747)	

† Military Temperature Range (-55° to 125°C)

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CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line
			Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL					
<b>Gate Arrays</b>														
													<b>(Cont'd)</b>	
4.0	38	CMOS												
			1740			64	x	x	x		3 to 10		691160	† IntIMicCir (4747)
			2160			100	x	x	x		3 to 10		671440	† IntIMicCir (4747)
			2646			84	x	x	x		3 to 10		691764	† IntIMicCir (4747)
			2940			116	x	x	x		3 to 10		671960	† IntIMicCir (4747)
			3360			68	x	x	x		3 to 10		692240	† IntIMicCir (4747)
			4185			84	x	x	x		3 to 10		692780	† IntIMicCir (4747)
			5250			106	x	x	x		3 to 10		693500	† IntIMicCir (4747)
			7650			100	x	x	x		3 to 10		695100	† IntIMicCir (4747)
4.2	38	CMOS	500	3		40	x	x	x		3 to 10		6C405	† Barven (4626)
			1000	3		65	x	x	x		3 to 10		6C410	† Barven (4626)
			1500	3		75	x	x	x		3 to 10		6C415	† Barven (4626)
			2000	3		90	x	x	x		3 to 10		6C420	† Barven (4626)
	55	CMOS	440			54		x	x		5		C440H	Fujitsu (4639)
			770			46		x	x		3 to 10		C770H	Fujitsu (4639)
			1275			64		x	x		5		C1275H	Fujitsu (4639)
			2000			72		x	x		5		C2000H	Fujitsu (4639)
			3900			72		x	x		5		C3900H	Fujitsu (4639, 4640)
	100	CMOS	1000			76	x	x	x		5, 2		CT0750	† Fairchild
4.4	30	CMOS	360			36	x	x	x		3 to 14		IS05A	† Siliconix
						36	x	x	x		3 to 14		IS05A	† Universal (4930)
			540			48	x	x	x		3 to 14		IS05B	† Siliconix
						48	x	x	x		3 to 14		IS05B	† Universal (4930)
			720			54	x	x	x		3 to 14		IS05C	† Siliconix
						54	x	x	x		3 to 14		IS05C	† Universal (4930)
			960			62	x	x	x		3 to 14		IS05D	† Siliconix
						62	x	x	x		3 to 14		IS05D	† Universal (4930)
			1200			68	x	x	x		3 to 14		IS05E	† Siliconix
						68	x	x	x		3 to 14		IS05E	† Universal (4930)
			1500			70	x	x	x		3 to 14		IS05F	† Siliconix
						70	x	x	x		3 to 14		IS05F	† Universal (4930)
			1800			82	x	x	x		3 to 14		IS05G	† Siliconix
						82	x	x	x		3 to 14		IS05G	† Universal (4930)
			2400			82	x	x	x		3 to 14		IS05H	† Siliconix
						82	x	x	x		3 to 14		IS05H	† Universal (4930)
80		CML	500			38	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA5RA	† Ferranti
						38	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA5RA	† Interdesign
			900			48	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA9RA	† Ferranti
						48	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA9RA	† Interdesign
			1200			52	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA12RA	† Ferranti
						52	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA12RA	† Interdesign
			1600			62	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA16RA	† Ferranti
						62	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA16RA	† Interdesign
			1800			64	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA18RA	† Ferranti
						64	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA18RA	† Interdesign
			2000			72	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA20RA	† Ferranti
						72	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA20RA	† Interdesign
			2400			80	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA24RA	† Ferranti
						80	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA24RA	† Interdesign
			3000			82	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA30RA	† Ferranti
						82	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA30RA	† Interdesign
			4000			118	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA40RA	† Ferranti
						118	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA40RA	† Interdesign

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line				
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL									
<b>Gate Arrays</b>															<b>(Cont'd)</b>				
4.5	25	CMOS	48		14	16	x	x	x		3 to 7		SP7000	† SPI	(4826)	5			
			80		16	20	x	x	x		2 to 7		SP7001	† SPI	(4826)				
			150		16	24	x	x	x		2 to 7		SP7002	† SPI	(4826)				
			300		22	44	x	x	x		2 to 7		SP7003	† SPI	(4826)				
					34	44	x	x	x		2 to 7		SP7004	† SPI	(4826)				
			544		30	56	x	x	x		2 to 7		SP7005	† SPI	(4826)				
			1000		40	76	x	x	x		2 to 7		SP7010	† SPI	(4826)				
					72	84	x	x	x		3 to 7		SP7210	SPI	(4826)				
			30	CMOS	400			40	x	x	x		5		CMA2004		† CMA	(4630)	10
					800			60	x	x	x		5		CMA2008		† CMA	(4630)	
1100					68	x	x	x		5		CMA2011	† CMA	(4630)					
1500					80	x	x	x		5		CMA2015	† CMA	(4630)					
2400					110	x	x	x		5		CMA2024	† CMA	(4630)					
70	TTL-LS	256	6		32		x	x		5	256 equiv. 3-input gates	μPB6101	NEC	(4804)	15				
		598	12		52		x	x			598 equiv. 3-input gates	μPB6102	NEC	(4804)					
80	CML	500			38	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA5RA	† Ferranti	20					
					38	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA5RA	† Interdesign						
		900		48	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA9RA	† Ferranti	25						
				48	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA9RA	† Interdesign							
		1200		52	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA12RA	† Ferranti	30						
				52	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA12RA	† Interdesign							
		1600		62	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA16RA	† Ferranti	35						
				62	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA16RA	† Interdesign							
		1800		64	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA18RA	† Ferranti	40						
				64	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA18RA	† Interdesign							
		2000		72	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA20RA	† Ferranti	45						
				72	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA20RA	† Interdesign							
		2400		80	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA24RA	† Ferranti	50						
				80	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA24RA	† Interdesign							
		3000		82	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA30RA	† Ferranti	55						
				82	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA30RA	† Interdesign							
4000		118	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA40RA	† Ferranti	60								
		118	x	x	x	x	3.5 to 5.5	300 μW/gate	ULA40RA	† Interdesign									
4.6	15	CMOS	504	(38)	(38)		x	x	x	3 to 6.5		MA5050	† Mitel	65					
			960	(48)	(48)		x	x	x	3 to 6.5		MA5100	† Mitel						
			1444	(62)	(62)		x	x	x	3 to 6.5		MA5150	† Mitel						
5.0	20	CMOS	252			38	x	x	x	3 to 10		SG250	† MicroTech	70					
			300		16	34	x	x	x	5 to 10, to 200	16 push-pull outputs, ± 15 mA, 200 V.	TMG5002	Telmos						
			405			40	x	x	x	3 to 10		SG400	† MicroTech						
			605			48	x	x	x	3 to 10		SG600	† MicroTech						
			798			54	x	x	x	3 to 10		SG800	† MicroTech						
			1012			58	x	x	x	3 to 10		SG1000	† MicroTech						
			1250			66	x	x	x	3 to 10		SG1250	† MicroTech						
			1510			72	x	x	x	3 to 10		SG1500	† MicroTech						
			1980			82	x	x	x	3 to 10		SG2000	† MicroTech						
			25	CMOS	880			66	x	x	x	3 to 6			LL5080	† LSILogic	(4762)	80	
					1404			84	x	x	x	3 to 6			LL5140	† LSILogic	(4762)		
					2224			106	x	x	x	3 to 6			LL5220	† LSILogic	(4762)		
					3192			130	x	x	x	3 to 6			LL5320	† LSILogic	(4762)		
					4202			148	x	x	x	3 to 6			LL5420	† LSILogic	(4762)		
6000					172	x	x	x	3 to 6		LL5600	† LSILogic	(4762)						
TTL-LS	320					28		x		5		TAL002	TI	(4887)					
	500					42		x		5		TAL004	TI	(4887)					

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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# IC MASTER

## CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL					
<b>Gate Arrays</b>															
<b>(Cont'd)</b>															
5.0	25	CMOS	840			40	x	x	x		3 to 7		CLA2100	† Plessey (4819)	
		CMOS	1440			52	x	x	x		3 to 7		CLA2300	† Plessey (4819)	
		CMOS	2400			64	x	x	x		3 to 7		CLA2500	† Plessey (4819)	
30		CMOS	150			32	x	x	x		3 to 14		AT-D-150	† ArrayTech	
			250	18	12	32	x	x	x		3 to 14		AT-D-250	† ArrayTech	5
			350			36	x	x	x		3 to 14		AT-D-350	† ArrayTech	
			550			48	x	x	x		3 to 14		AT-D-550	† ArrayTech	
			750			54	x	x	x		3 to 14		AT-D-750	† ArrayTech	
			1000			62	x	x	x		3 to 14		AT-D-1000	† ArrayTech	
			1200			68	x	x	x		3 to 14		AT-D-1200	† ArrayTech	10
			1500			76	x	x	x		3 to 14		AT-D-1500	† ArrayTech	
			1800			82	x	x	x		3 to 14		AT-D-1800	† ArrayTech	
33		CMOS	500	3		40	x	x	x		3 to 10		BC405	† Barvon	
			1000	3		65	x	x	x		3 to 10		BC410	† Barvon	
			1500	3		75	x	x	x		3 to 10		BC415	† Barvon	15
			2000	3		90	x	x	x		3 to 10		BC420	† Barvon	
40		CMOS	725			58	x				5		PA40650	† RCA (4824)	
			953			70	x				5		PA40850	† RCA (4824)	
			1133			78	x				5		PA41000	† RCA (4824)	
			1313			86	x				5		PA41200	† RCA (4824)	20
50		CMOS	600			38	x	x	x		5		7CL	† STC	
			648	2		35	x	x	x		3 to 6		HCA6306	† Motorola	
			725			58	x				4 to 10		PA60650	† RCA (4824)	
			1200	17		42	x	x	x		3 to 6		HCA6312	† Motorola	
			1313			86	x				4 to 10		PA61200	† RCA (4824)	25
			1440			64	x	x	x		5		7CM	† STC	
			1920			96	x	x	x		5		7CK	† STC	
			2295	55		56	x	x	x		3 to 6		HCA6324	† Motorola	
			4200			80	x	x	x		3 to 7		CLA3000	† Plessey (4819)	
			4860	53		54	x	x	x		3 to 6		HCA6348	† Motorola	30
5.0 *	15	CMOS	504			38	x	x	x		3 to 6.5		50500A	GTEMicro	
			960			48	x	x	x		3 to 6.5		51000A	GTEMicro	
			1440			62	x	x	x		3 to 6.5		51500A	GTEMicro	
6 *	25	CMOS	1200			64	x	x	x		5		TAC020	TI	
6.0	12	CMOS	408			34	x	x	x		3 to 9		IGC10408	† Intersil (4751)	35
			756			44	x	x	x		3 to 9		IGC10756	† Intersil (4751)	
			1500			62	x	x	x		3 to 9		IGC11500	† Intersil (4751)	
			2001			70	x	x	x		3 to 9		IGC12001	† Intersil (4751)	
15		CMOS	324			30	x	x	x		3 to 6		LAD3	† GI	
			560			38	x	x	x		3 to 6		LA05	† GI	40
			960			50	x	x	x		3 to 6		LA10	† GI	
			1440			62	x	x	x		3 to 6		LA15	† GI	
			2014			74	x	x	x		3 to 6		LA20	† GI	
20		ISL	1192			36		x	x		5, 1.5		8A1200	† Signetics	
						60		x	x		5, 1.5		8A1260	† Signetics	45
			1472			42		x	x		5, 1.5		8A1542	† Signetics	
			1620			64		x	x		5, 1.5		8A1664	† Signetics	
			1740			64		x	x		5, 1.5		8A1864	† Signetics	
			2088			76		x	x		5, 1.5		8A2176	† Signetics	
25		CMOS	427	2		34	x	x	x		5		μPD65003	NEC (4809)	50
			500	3		40	x	x	x		3 to 10		BC505	† Barvon (4826)	
			858			48	x	x	x		5		μPD65002	NEC (4809)	
			1000	3		65	x	x	x		3 to 10		BC510	Barvon (4826)	
			1368			62	x	x	x		5		μPD65010	NEC (4809)	
			1500	3		75	x	x	x		3 to 10		BC515	† Barvon (4826)	55

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line	
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL						
<b>Gate Arrays</b>															<b>(Cont'd)</b>	
6.0	25	CMOS	2000	3		90	x	x	x		3 to 10		BC520	† Barvon	(4626)	
			2112			78	x	x	x		5		μPD65020	NEC	(4809)	
6.5 *	—	CMOS	200			26		x	x		5	208 equiv. 3-input gates	B200	Fujitsu	(4637)	
6.7	50	CMOS	640			38	x	x	x		5		MKA300	† Mostek	5	
			2048			92	x	x	x		5		MKA102A	† Mostek		
			3680			130	x	x	x		5		MKA1840	† Mostek		
			5376			138	x	x	x		5		MKA2688	† Mostek		
7.0	10	CMOS	1250			80	x	x	x		5	XRCSD	† Exar			
7.2	27	CMOS	500	3		40	x	x	x		3 to 10		BC505	† Barvon	10	
			1000	3		65	x	x	x		3 to 10		BC520	† Barvon		
			1500	3		75	x	x	x		3 to 10		BC515	† Barvon		
			2000	3		90	x	x	x		3 to 10		BC520	† Barvon		
8.0	10	NMOS	370	2	16					x	5	Logic array	WD1820	Western	15	
			1760	3		35					x	5	Logic array with PLA	WD1828/40		Western
8.0	20	CMOS	168			20	x				3 to 12		MA10150	† RCA	(4824)	20
			182			20	x				4 to 10		MA60150	† RCA	(4824)	
			276			24	x				3 to 12		MA16250	† RCA	(4824)	
			300			24	x						MA60250	† RCA	(4824)	
			317			30	x				5		PA20254	† RCA	(4824)	
			400			48	x				3 to 12		MA10350	† RCA	(4824)	
			452			32	x				4 to 10		MA60400	† RCA	(4824)	
			517			46	x				5		PA20450	† RCA	(4824)	
			576			64	x				3 to 12		MA10500	† RCA	(4824)	
			632			32	x				4 to 10		MA60550	† RCA	(4824)	
			725			58	x				5		PA30650	† RCA	(4824)	
			848			32	x				4 to 10		MA60800	† RCA	(4824)	
			953			70	x				5		PA20850	† RCA	(4824)	
			1313			78	x				5		PA21000	† RCA	(4824)	
8.0	33	CMOS	300			38	x	x	x		1 to 12		TM4030	† Telmos	25	
			338	8		37	x	x	x		5 to 10, to 300	40 dedicated flip-flops, 8 300V transistors	TM5001	† Telmos		
			400			44	x	x	x		1 to 12		TM4040	† Telmos		
			540			50	x	x	x		1 to 12		TM4054	† Telmos		
			770			60	x	x	x		1 to 12		TM4077	† Telmos		
			1000			68	x	x	x		1 to 12		TM4100	† Telmos		
			1260			76	x	x	x		1 to 12		TM4126	† Telmos		
			40	CMOS	150			29	x	x	x		2 to 7			MPE
510			54		x	x	x		2 to 7		MPA	† Interdesign (4746)				
1000			73		x	x	x		2 to 7		MPB	† Interdesign (4746)				
1500			89		x	x	x		2 to 7		MPC	† Interdesign (4746)				
2000			122	x	x	x		2 to 7		MPP	† Interdesign (4746,4747)					
8.4	35	CMOS	770			46	x	x	x		5		C770	Fujitsu	40	
			1275			64	x	x	x		5		C1275	Fujitsu		
			2000			72	x	x	x		5		C2000	Fujitsu		
			3900			72	x	x	x		5		C3900	Fujitsu (4640)		
9.0	15	CMOS	272			32	x	x	x		3 to 6		LL3020	† LSI Logic (4762)	50	
			342			36	x	x	x		3 to 6		LL3030	† LSI Logic (4762)		
			420			40	x	x	x		3 to 6		LL3040	† LSI Logic (4762)		
			600			48	x	x	x		3 to 6		LL3060	† LSI Logic (4762)		
			812			56	x	x	x		3 to 6		LL3080	† LSI Logic (4762)		
			1056			64	x	x	x		3 to 6		LL3110	† LSI Logic (4762)		
			1332			72	x	x	x		3 to 6		LL3130	† LSI Logic (4762)		
			1722			82	x	x	x		3 to 6		LL3170	† LSI Logic (4762)		
			2162			92	x	x	x		3 to 6		LL3210	† LSI Logic (4762)		
			2550			100	x	x	x		3 to 6		LL3250	† LSI Logic (4762)		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line	
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL						
<b>Gate Arrays</b>																
													<b>(Cont'd)</b>			
10	8	CMOS	153			30	x	x	x		1.5 to 18	4 dedicated flip-flops	HI3600	† Holt	(4745)	
			384			44	x	x	x		1.5 to 18	16 dedicated flip-flops	HI3900	† Holt	(4745)	
			490			48	x	x	x		1.5 to 18	32 dedicated flip-flops	HI3100	† Holt	(4745)	
			1020			68	x	x	x		1.5 to 18	60 dedicated flip-flops	HI3300	† Holt	(4745)	
10	CMOS	300	1		37	x	x	x		2.7 to 12		UA-1	† AMI	(4619)	5	
		400	1		43	x	x	x		2.7 to 12		UA-2	† AMI	(4619)		
		540	1		49	x	x	x		2.7 to 12		UA-3	† AMI	(4619)		
		770	1		59	x	x	x		2.7 to 12		UA-4	† AMI	(4619)		
		1000	1		67	x	x	x		2.7 to 12		UA-5	† AMI	(4619)		
		1260	1		75	x	x	x		2.7 to 12		UA-6	† AMI	(4619)	10	
10.0	20	CMOS	560			38	x	x	x		3 to 7		CLA1000	† Plessey	(4819)	15
			964			50	x	x	x		3 to 7		CLA1200	† Plessey	(4819)	
			1400			62	x	x	x		3 to 7		CLA1500	† Plessey	(4819)	
			12	6	CMOS	300	1		37	x	x	x		3 to 10		LC3100
400	1		43			x	x	x		3 to 10		LC4100	† LSI Logic	(4762)		
540	1		49			x	x	x		3 to 10		LC5400	† LSI Logic	(4762)		
770	1		59			x	x	x		3 to 10		LC7700	† LSI Logic	(4762)		
1000	1		67			x	x	x		3 to 10		LC10000	† LSI Logic	(4762)		
1260	1		75			x	x	x		3 to 10		LC12600	† LSI Logic	(4762)		
12.7	20	CML	2000			64	x	x	x	x	3.5 to 5.5	250 μW/gate	ULA9C	† Ferranti		25
						64	x	x	x	x	3.5 to 5.5	250 μW/gate	ULA9C	† Interdesign		
			2400			68	x	x	x	x	3.5 to 5.5	250 μW/gate	ULA12C	† Ferranti		30
						68	x	x	x	x	3.5 to 5.5	250 μW/gate	ULA12C	† Interdesign		
			500			38	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA5RB	† Ferranti		35
						38	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA5RB	† Interdesign		
			900			48	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA9RB	† Ferranti		40
						48	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA9RB	† Interdesign		
			1200			52	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA12RB	† Ferranti		45
						52	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA12RB	† Interdesign		
1600			62	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA16RB	† Ferranti		50			
			62	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA16RB	† Interdesign					
1800			64	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA18RB	† Ferranti		55			
			64	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA18RB	† Interdesign					
2000			72	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA20RB	† Ferranti		55			
			72	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA20RB	† Interdesign					
2400			80	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA24RB	† Ferranti		55			
			80	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA24RB	† Interdesign					
3000			82	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA30RB	† Ferranti		55			
			82	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA30RB	† Interdesign					
4000			118	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA40RB	† Ferranti		55			
			118	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA40RB	† Interdesign					
13	5	CMOS	50			22	x	x	x		3 to 15		TM3050	† Telmos		
			112			32	x	x	x		3 to 15		TM3100	† Telmos		
			162			38	x	x	x		3 to 15		TH3150	† Telmos		
			216			44	x	x	x		3 to 15		TM3200	† Telmos		
			224			53	x	x	x		3 to 15	32 dedicated flip-flops	TM3350	† Telmos	50	
			272			69	x	x	x		3 to 15	72 dedicated flip-flops	TM3600	† Telmos		
464			69	x	x	x		3 to 15	24 dedicated flip-flops	TM3500	† Telmos					
13.6	20	CML	500			38	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA5RB	† Ferranti		
						38	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA5RB	† Interdesign		
			900			48	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA9RB	† Ferranti		
						48	x	x	x	x	3.5 to 5.5	100 μW/gate	ULA9RB	† Interdesign		

(Continued)

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line				
			Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL									
<b>Gate Arrays</b>														(Cont'd)				
13.6	20	CML	1200		52	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA12RB	† Ferranti	5				
					52	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA12RB	† Interdesign					
			1600		62	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA16RB	† Ferranti					
					62	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA16RB	† Interdesign					
			1800		64	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA18RB	† Ferranti					
					64	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA18RB	† Interdesign					
			2000		72	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA20RB	† Ferranti					
					72	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA20RB	† Interdesign					
			2400		80	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA24RB	† Ferranti					
					80	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA24RB	† Interdesign					
			3000		82	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA30RB	† Ferranti					
					82	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA30RB	† Interdesign					
			4000		118	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA40RB	† Ferranti					
					118	x	x	x	x	3.5 to 5.5	100 $\mu$ W/gate	ULA40RB	† Interdesign					
			14	20	CML	450		39	x	x	x	x	3.5 to 5.5		250 $\mu$ W/gate	ULA2C	† Ferranti	15
								39	x	x	x	x	3.5 to 5.5		250 $\mu$ W/gate	ULA2C	† Interdesign	
900		48				x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA5C	† Ferranti					
		48				x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA5C	† Interdesign					
15	1	PL	1120		96	x	x	x			SBP96700	† TI	(538)	20				
			2120		140	x	x	x			SBP96600	† TI	(538)					
	2	CMOS	80		28	x	x	x		3 to 15		606	† MicroTech	25				
			160		38	x	x	x		3 to 15		616	† MicroTech					
			320		48	x	x	x		3 to 15		626	† MicroTech					
			480		62	x	x	x		3 to 15		636	† MicroTech					
			1000		72	x	x	x		3 to 15		646	† MicroTech					
			2000		120				x	5 to 7	12 dedicated flip-flops		707		† MicroTech			
	20	CML	454		39	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA2C	† Ferranti	30				
				39	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA2C	† Interdesign						
			900		48	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA5C	† Ferranti					
				48	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA5C	† Interdesign						
	2000		64	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA9C	† Ferranti	35						
			64	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA9C	† Interdesign							
		2400		68	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA12C		† Ferranti					
			68	x	x	x	x	3.5 to 5.5	250 $\mu$ W/gate	ULA12C	† Interdesign							
16	6*	CMOS	330		38	x	x	x		3 to 15		SCC0330N	Signetics (4862)	40				
			448		26	x	x	x		3 to 15		SCC0450N	Signetics (4862)					
			704		38	x	x	x		3 to 15		SCC0700N	Signetics (4862)					
			1106		66	x	x	x		3 to 15		SCC1100N	Signetics (4862)					
	10	CMOS	168		20	x				1.1 to 20		MA30150	† RCA					
			276		24	x						MA30250	† RCA					
			140		29	x	x	x		3 to 15		MCA	† Interdesign (4746)					
			200		34	x	x	x		3 to 15		MCB	† Interdesign (4746)					
17	9	CMOS	270		40	x	x	x		3 to 15		MCC	† Interdesign (4746)	45				
			340		40	x	x	x		3 to 15	28 dedicated flip-flops	MCE	† Interdesign (4746)					
			440		46	x	x	x		3 to 15	32 dedicated flip-flops	MCD	† Interdesign (4746)					
			630		50	x	x	x		3 to 15		MCF	† Interdesign (4746)					
			880		58	x	x	x		3 to 15		MCS	† Interdesign (4746)					
			18	3	CMOS	100		32	x	x		0.5 to 18			Interchip A	† Intech	50	
150		38	x			x			0.5 to 18		Interchip B	† Intech						
215		44	x			x			0.5 to 18		Interchip C	† Intech						
224		53	x			x			0.5 to 18	32 dedicated flip-flops	Interchip D	† Intech						
368		64	x			x			0.5 to 18	72 dedicated flip-flops	Interchip F	† Intech						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line		
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL							
<b>Gate Arrays</b>															<b>(Cont'd)</b>		
18	3	CMOS	464			64	x	x			0.5 to 18	24 dedicated flip-flops	Interchip E	† Intech	(Cont'd)		
21	2.5	CMOS	112			29	x	x	x		3 to 12		XRCMA	† Exar	5		
			162			34	x	x	x		3 to 12		XRCMB	† Exar			
			216			40	x	x	x		3 to 12		XRCMC	† Exar			
			544			46	x	x	x		3 to 12		XRCMD	† Exar			
25	8	CMOS	100	11		18	x	x	x		3 to 18		CDI1000	† CalDevices	10		
			200	12		26	x	x	x		3 to 18		CDI2000	† CalDevices			
			315	14		32	x	x	x		3 to 18		CDI3000	† CalDevices			
			420	14		38	x	x	x		3 to 18		CDI4000	† CalDevices			
			600	18		44	x	x	x		3 to 18		CDI6000	† CalDevices			
			880	16		54	x	x	x		3 to 18		CDI8800	† CalDevices			
			12	CMOS	220	8	10		x	x	x		1 to 18			MGA220	† MCE
					225	8	10		x	x	x		1 to 18			MGA255	† MCE
30	10	1 <sup>†</sup> L	11400			62			x	x	1 to 6	Programmable speed/power, 7600 equiv. 5-input gates	GAC-8000	† CIC			
42	10	CML	500			38	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA5RC	† Ferranti	15		
						38	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA5RC	† Interdesign			
			900			48	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA9RC	† Ferranti	20		
						48	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA9RC	† Interdesign			
			1200			52	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA12RC	† Ferranti	25		
						52	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA12RC	† Interdesign			
			1600			62	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA16RC	† Ferranti	30		
						62	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA16RC	† Interdesign			
			1800			64	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA18RC	† Ferranti	35		
						64	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA18RC	† Interdesign			
			2000			72	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA20RC	† Ferranti	40		
						72	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA20RC	† Interdesign			
			2400			80	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA24RC	† Ferranti	45		
						80	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA24RC	† Interdesign			
			3000			82	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA30RC	† Ferranti	50		
						82	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA30RC	† Interdesign			
4000			118	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA40RC	† Ferranti						
			118	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA40RC	† Interdesign						
45	5	CML	450			39	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA2N	† Ferranti	35		
						39	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA2N	† Interdesign			
			900			48	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA5N	† Ferranti	40		
						48	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA5N	† Interdesign			
			2000			64	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA9N	† Ferranti	45		
						64	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA9N	† Interdesign			
			2400			68	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA12N	† Ferranti	50		
						68	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA12N	† Interdesign			
10	CML	500			38	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA5RC	† Ferranti	55			
					38	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA5RC	† Interdesign				
		900			48	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA9RC	† Ferranti	60			
					48	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA9RC	† Interdesign				
		1200			52	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA12RC	† Ferranti	65			
					52	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA12RC	† Interdesign				
		1600			62	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA16RC	† Ferranti	70			
					62	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA16RC	† Interdesign				
1800			64	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA18RC	† Ferranti	75					
			64	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA18RC	† Interdesign						
2000			72	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA20RC	† Ferranti	80					
			72	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA20RC	† Interdesign						
2400			80	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA24RC	† Ferranti						
			80	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA24RC	† Interdesign						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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CUSTOM/SEMICUSTOM

CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line	
			Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL						
<b>Gate Arrays</b>														<b>(Cont'd)</b>	
45	10	CML	3000			82	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA30RC	† Ferranti	(Cont'd)
						82	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA30RC	† Interdesign	
			4000			118	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA40RC	† Ferranti	
						118	x	x	x	x	3.5 to 5.5	30 μW/gate	ULA40RC	† Interdesign	
48	5	CML	450			39	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA2N	† Ferranti	5
						39	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA2N	† Interdesign	
			900			48	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA5N	† Ferranti	
						48	x	x	x	x	3.5 to 5.5	70 μW/gate	ULA5N	† Interdesign	
			2000			64	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA9N	† Ferranti	
						64	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA9N	† Interdesign	10
			2400			68	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA12N	† Ferranti	
						68	x	x	x	x	3.5 to 5.5	40 μW/gate	ULA12N	† Interdesign	
50	2	CMOS	75			23	x	x	x		2 to 15		G4060	† Int!MicCir	
			140			29	x	x	x		2 to 15		G4112	† Int!MicCir	
			200			35	x	x	x		2 to 15		G4160	† Int!MicCir (4747)	15
			256			18	x	x	x		0.1 to 100 μA	Programmable speed/power	D15C	† MCE	
			275			38	x	x	x		2 to 15		G4220	† Int!MicCir (4747)	
			288			28	x	x	x		0.1 to 100 μA	Programmable speed/power	D15A	† MCE	
			330			41	x	x	x		2 to 15		G4264	† Int!MicCir (4747)	
			390			43	x	x	x		2 to 15		G4312	† Int!MicCir (4747)	20
			455			47	x	x	x		2 to 15		G4364	† Int!MicCir (4747)	
			520			40	x	x	x		0.1 to 100 μA	Programmable speed/power	D15B	† MCE	
			525			49	x	x	x		2 to 15		G4420	† Int!MicCir (4747)	
			600			53	x	x	x		2 to 15		G4480	† Int!MicCir (4747)	
			864			21	x	x	x		0.1 to 100 μA	Programmable speed/power	D15D	† MCE	25
		FL	192			24	x	x	x	x	1 to 12	Programmable speed/power	1200	Cherry	
						24	x	x	x		1.5 to 7	Programmable speed/power	XR200	† Exar (3386)	
			288			28	x	x	x	x	1 to 12	Programmable speed/power	1300	Cherry	
						28	x	x	x		1.5 to 7	Programmable speed/power	XR300	† Exar (3386)	30
			520			40	x	x	x		1.5 to 7	Programmable speed/power	XR500	† Exar (3386)	
140	5	CML	500			38	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA5RD	† Ferranti	
						38	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA5RD	† Interdesign	
			900			48	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA9RD	† Ferranti	
						48	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA9RD	† Interdesign	
			1200			52	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA12RD	† Ferranti	35
						52	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA12RD	† Interdesign	
			1600			62	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA16RD	† Ferranti	
						62	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA16RD	† Interdesign	
			1800			64	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA18RD	† Ferranti	
						64	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA18RD	† Interdesign	40
			2000			72	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA20RD	† Ferranti	
						72	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA20RD	† Interdesign	
			2400			80	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA24RD	† Ferranti	
						80	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA24RD	† Interdesign	
			3000			82	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA30RD	† Ferranti	45
						82	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA30RD	† Interdesign	
			4000			118	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA40RD	† Ferranti	
						118	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA40RD	† Interdesign	
220	5	CML	500			38	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA5RD	† Ferranti	50
						38	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA5RD	† Interdesign	
			900			48	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA9RD	† Ferranti	
						48	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA9RD	† Interdesign	
			1200			52	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA12RD	† Ferranti	
						52	x	x	x	x	3.5 to 5.5	12 μW/gate	ULA12RD	† Interdesign	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.



# IC MASTER

## CUSTOM/SEMICUSTOM-Gate Arrays (Cont'd)

Internal Gate Delay ns (max)	Toggle Frequency MHz	Type	No. Equiv. 2-Input Gates	Input/Output Cells			Input/Output Compatibility				Supply Voltage V	Comments	Device	Source	Line
				Input Only	Output Only	Un-committed	CMOS	TTL-LS	TTL	ECL					
<b>Gate Arrays</b>															
<b>(Cont'd)</b>															
220	5	CML	1600			62	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA16RD	† Ferranti	5
						62	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA16RD	† Interdesign	
				1800		64	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA18RD	† Ferranti	
						64	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA18RD	† Interdesign	
				2000		72	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA20RD	† Ferranti	
						72	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA20RD	† Interdesign	
			2400		80	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA24RD	† Ferranti		
					80	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA24RD	† Interdesign		
			3000		82	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA30RD	† Ferranti		
					82	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA30RD	† Interdesign		
			4000		118	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA40RD	† Ferranti		
					118	x	x	x	x	3.5 to 5.5	12 $\mu$ W/gate	ULA40RD	† Interdesign		
8000	0.1	CML	510			40	x	x	x	x	3.5 to 5.5	0.4 $\mu$ W/gate	ULA2M	† Ferranti	10
						40	x	x	x	x	3.5 to 5.5	0.4 $\mu$ W/gate	ULA2M	† Interdesign	
10500	0.1	CML	510			40	x	x	x	x	3.5 to 5.5	0.4 $\mu$ W/gate	ULA2M	† Ferranti	15
						40	x	x	x	x	3.5 to 5.5	0.4 $\mu$ W/gate	ULA2M	† Interdesign	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

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CUSTOM/SEMICUSTOM-Linear & Linear/Digital Arrays

Manufacturer Technology	Total Bonding Pads	Linear		Digital		Uncommitted Transistors		Number Capacitors	Supply Voltage, V	Comments	Device	Source	Line
		Dedicated Op Amps, A/D, etc.	Total Other Lin. Components	Equiv. 2-Input Gates	Gate Delay ns @ V	NPN	PNP						
<b>Linear &amp; Linear/Digital Arrays</b>													
Array Technology CMOS	42	6		200	5	500	500	12	12	8-bit DAC, 4 op amps, 2 comparators	AT-LD-04	ArrayTech	
	70	12		400	5	1000	1000	24	12	8-bit DAC, 8 op amps, 4 comparators	AT-LD-08	ArrayTech	
Barvon Research Bi-CMOS		1	2292	400	7	104	76		15/6	4 to 16 decades, combination bipolar and CMOS	BC3602	Barvon (4626)	
	58	—	585	250	7	92	52		15/6	Combination bipolar and CMOS	BC3601	Barvon (4626)	
California Devices BIP	14		110			22	8		20		CLIC-C	CalDevices	5
	16		209			50	16		20		CLIC-D	CalDevices	
			260			59	15		20		CLIC-A	CalDevices	
	18		168			38	12		20		CLIC-I	CalDevices	
			207			48	15		20		CLIC-E	CalDevices	
			460			96	36		20		CLIC-F	CalDevices	10
	20		370			72	22		20		CLIC-H	CalDevices	
	24		300			69	12		20		CLIC-B	CalDevices	
			310			60	18		20		CLIC-G	CalDevices	
			398			80	22		20		CLIC-L	CalDevices	
Cherry Semiconductor BIP	16		333			60	25		20	Flip-chip	CS2800	Cherry	15
	18		187			48	15		20		CS2000E	Cherry	
			325			60	18		20		CS2500G	Cherry	
	22		299			61	24		20	Photosensor on-chip	CS3500	Cherry	
			479			88	36		20		CS3200L	Cherry	
	24		437			96	36		20		CS3000F	Cherry	20
	25		695			123	52		20		CS3600	Cherry	
	28		801			153	56		20		CS4000	Cherry	
PL	40		300	256					12	69 NPN/PNP transistors, ion implant resistors	CML1400	Cherry	
			300	256					12	69 NPN/PNP transistors, 18 input/output cells	CS1400	Cherry	
Custom Integrated Circuits BIP	10		38			25	8	33	20	1 cell, often used in multiples, PNPs have 4 collectors	MSA-1	CIC	25
	38		124			84	40	124	20	1 cell can be used in multiples	MSB-1	CIC	
	50		800			561	164	725	20	15 cells	MSA-15	CIC	
	56		148			168	80	148	20	2 cells	MSB-2	CIC	
Exar BIP	14		110			23	8		20		XR-C100	Exar (3386)	
	16		209			50	16		36		XR-D100	Exar (3386)	30
			260			60	18		20		XR-A100	Exar (3386)	
			300			69	12		20		XR-B100	Exar (3386)	
	18		200			48	15		20		XR-E100	Exar (3386)	
			230			34	16		75	High voltage array	XR-X100	Exar	
			260			38	12		20		XR-J100	Exar (3386)	35
			309			58	18		20		XR-G100	Exar (3386)	
			401			75	22		20		XR-H100	Exar (3386)	
	24		251			80	26		20		XR-L100	Exar (3386)	
			472			93	36		20		XR-F100	Exar (3386)	
	28		539			128	52	8	36	8 JFETs, ion implant resistors	XR-V100	Exar (3386)	40
			812			148	56		20		XR-M100	Exar (3386)	
	40		901			208	68		36	8 JFETs, ion implant resistors	XR-W100	Exar (3386)	
PL	40		401	256		45	12		12	6 or 12V process	XR400	Exar (3386)	
Ferranti Electric CML/RTL	25		908	100		407		2	5.5	Precision reference, shaping capacitors, 100 mA transistors, 1V operation	ULA1G	Ferranti	
	28		1080	150		384			5.5	1V operation	ULA1L	Ferranti	

\* Delay conditions: 2-Input NAND driving 2 NANDs (F0 = 2), 5V supply, 70°C, worst-case processing.

(Continued)

CUSTOM/SEMICUSTOM-Linear & Linear/Digital Arrays (Cont'd)

Manufacturer	Total Bonding Pads	Linear		Digital		Uncommitted Transistors		Number Capacitors	Supply Voltage, V	Comments	Device	Source	Line
		Dedicated Op Amps, A/D, etc.	Total Other Lin. Components	Equiv. 2-Input Gates	Gate Delay ns @ V	NPN	PNP						
<b>Linear &amp; Linear/Digital Arrays</b>													<b>(Cont'd)</b>
Ferranti Electric CML/RTL	28		1390	160		647		2	5.5	Precision reference, shaping capacitors, 100 mA transistors, 1V operation	ULA2G	Ferranti	(Cont'd)
	30		1327	280		676			5.5	1V operation	ULA1U	Ferranti	
	40		1920	340		715			5.5	1V operation	ULA2L	Ferranti	
			2377	510		1280			5.5	1V operation	ULA2U	Ferranti	
			2781	580		1589			5.5	1V operation	ULA3U	Ferranti	5
	48		4304	730		1644			5.5	1V operation	ULA5L	Ferranti	
Holt CMOS	52	46		36				105	15	8-bit DAC, 11 op amps, 1 comparator, 22 flip-flops, 4 switched-capacitor filters	HI3700	Holt	(4745)
Integrated Circuits BIP	14					22	8		20		IMC-20C	IntCirSys	
	16					59	18		20		IMC-20A	IntCirSys	10
						50	16		40		IMC-400	IntCirSys	
						62	13		40		IMC-40B	IntCirSys	
	18					48	0		20		IMC-20E	IntCirSys	
						60	0		20		IMC-20G	IntCirSys	
						72	22		20		IMC-20H	IntCirSys	
						38	0		20		IMC-20J	IntCirSys	15
	20					60	28		40		IMC-20B	IntCirSys	
	24					96	0		20		IMC-20F	IntCirSys	
						69	12		20		IMC-20L	IntCirSys	
						80	0		20		IMC-40A	IntCirSys	
						96	14		20		IMC	IntCirSys	20
Interdesign BIP	14		110			22	8		20		MOC	Interdesign	(4746)
	16		194			50	16		36		MOD	Interdesign	(4746)
			260			59	18		20		MOA	Interdesign	(4746)
	18		170			38	12		20		MOJ	Interdesign	(4746)
			187			48	15		20		MOE	Interdesign	(4746)
			314			56	22		20		MOQ	Interdesign	(4746)
			382			72	22		20		MOH	Interdesign	(4746)
	20		310			60	18		20		MOB	Interdesign	(4746)
	24		300			69	12		20		MOB	Interdesign	(4746)
			416			80	26		20		MOL	Interdesign	(4746)
			437			96	36		20		MOF	Interdesign	(4746)
			571			72	57		20		MOP	Interdesign	(4746)
	28		812			141	56		20		MOM	Interdesign	(4746)
	40		1072			182	70		20		MON	Interdesign	(4746)
CDI	40		2000	265		715			5.5		MUA	Interdesign	(4746)
			2000	265		715			5.5		MUB	Interdesign	(4746)
			2000	265		715			5.5		MUC	Interdesign	(4746)
CMOS	42		930	199		414	398	84	15	16 dedicated flip-flops, switched capacitor array	MLA	Interdesign	(4746)
MCE Semiconductor BIP	14					16	18	2	20	Power array	MCEP20A	MCE	
			109			22	8		20		MCEA20C	MCE	40
			109			62	8		20		MCEA20CS	MCE	
	16		194			100	16		40		MCEA40D	MCE	
			194			50	16		40		MCEA40DS	MCE	
			245			60	28		40		MCEA40AS	MCE	
			259			59	18		20		MCEA20A	MCE	45
			259			59	18		20		MCEA20AS	MCE	
	18		170			38	12		20		MCEA20J	MCE	

\* Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

Master Selection Guide

CUSTOM/SEMICUSTOM

CUSTOM/SEMICUSTOM-Linear & Linear/Digital Arrays (Cont'd)

Manufacturer	Total Bonding Pads	Linear		Digital		Uncommitted Transistors		Number Capacitors	Supply Voltage, V	Comments	Device	Source	Line
		Dedicated Op Amps, A/D, etc.	Total Other Lin. Components	Equiv. 2-Input Gates	Gate Delay ns @ V	NPN	PNP						
<b>Linear &amp; Linear/Digital Arrays</b>													(Cont'd)
MCE Semiconductor	18												(Cont'd)
			170			38	12		20		MCEA20JS	MCE	
			187			48	15		20		MCEA20E	MCE	
			187			48	15		20		MCEA20ES	MCE	
			308			60	18		20		MCEA20G	MCE	
			308			60	18		20		MCEA20GS	MCE	5
			382			72	22		20		MCEA20H	MCE	
			382			74	22		20		MCEA20HS	MCE	
	20		—			27	12		40	5 op amps, power array	MCEP40A	MCE	
			218			62	10		40		MCEA40BS	MCE	
	24		301			69	12		20		MCEA20B	MCE	10
			301			69	12		20		MCEA20BS	MCE	
			416			80	26		20		MCEA20L	MCE	
			416			80	26		20		MCEA20LS	MCE	
			437			96	36		20		MCEA20F	MCE	
			437			96	36		20		MCEA20FS	MCE	15
			464			96	36		20		MCEA20WS	MCE	
	28					149	56		20		MCEA20M	MCE	
			812			149	56		20		MCEA20MS	MCE	
CMOS	18	2						2	8	2 op amps, dedicated logic	MGA255	† MCE	
		2						2	12	2 op amps, dedicated logic	MGA255	† MCE	20
		2				8 @ 15			2	18	MGA255	† MCE	
Plessey Semiconductor	24												
			278			81	28		24		BAA1000	Plessey	
			572			159	58		24		BAA2000	Plessey	
Reticon	60	31						2500	± 11	14 filter sections, 3 extra op amps, 20 state cells, min. package size 25 pins, switched capacitor filter array	R5626	Reticon	
Silicon Systems Inc.	28	24						396	15	12 op amps, 210 switches, 4 buffers, 8 flip-flops, switched capacitor filter array	SCA-6	SiliconSys (3685.4872)	25
	54	46						720	12	24 op amps, 456 switches, 6 buffers, 16 flip-flops, switched capacitor filter array	SCA-12	SiliconSys (3685.4872)	
STC Microtechnology	18		274			64	10		15	Ft=1.5 GHz	MSIV-1	STC	
	26		548			128	20		15	Ft=1.5 GHz	MSIV-2	STC	
	34		871			258	32	2	15		MSII	STC	
	40		1056			256	40		15	Ft=1.5 GHz	MSIV-4	STC	30
Telmos	36			290	5			54	10	16 dedicated flip-flops, 8 op amps, 36 switches	TM6005	Telmos (4877)	
	42			530	5			80	10	32 dedicated flip-flops, 16 op amps, 32 switches	TM6001	Telmos (4877)	
	46			320	5			452	10	37 op amps, 100 switches	TM6003	Telmos (4877)	
	55			740	5			232	10	40 dedicated flip-flops, 16 op amps, 100 switches	TM6002	Telmos (4877)	
	62			800	5			244	10	32 dedicated flip-flops, 22 op amps, 104 switches	TM6004	Telmos (4877)	35

\* Delay conditions: 2-Input NAND driving 2 NANDs (FO = 2), 5V supply, 70°C, worst-case processing.

CUSTOM/SEMICUSTOM-Standard Cells

Manufacturer Device Process Technology	Geometry (μm)	Interconnections		Delay 2-Input NAND Gate (ns)	Cell Library Includes						Supply Voltage, V	Comments	Library or Process Name	Source	Line
		Metal	Poly		A/D	D/A	Op Amp	Core μP	RAM	ROM					
<b>Standard Cells</b>															
Alphatron CMOS, Si-Gate	3	1	1				x		x	x		3 to 6		ALPHAMAP † Alphatron	
	4	1	1				x		x	x		3 to 6	74LS MSI speeds	ALPHAMAP † Alphatron	
American Microsystems CMOS, Si-Gate	3	1	1	2.5								3 to 10		CMOS Library † AMI (4624)	
		2	1	2				x				3 to 10		CMOS Library † AMI (4624)	
	5	1	0	7				x				3 to 15		CMOS Library † AMI (4624)	5
NMOS, Si-Gate	4	1	0	7								5		NMOS Library † AMI (4624)	
Array Technology CMOS, Si-Gate	3	1	1	4		x	x	x	x	x		±5		CMOS Library † ArrayTech	
	NMOS, Si-Gate	4	1	1	6				x	x	x	3 to 7		NMOS Library † ArrayTech	
Barvon CMOS, Si-Gate	3	2		1.8	x	x	x	x	x	x	x	3 to 10	Core μP, 1084	BHC Library Barvon (4626)	
	HMOS	3	2										Avail. 1084	BHN Library Barvon (4626)	10
Circuit Technology CMOS, Si-Gate	4	1	1	5			x					3 to 10	200°C operation	CELLMOS † CirTech	
	5	1	1	6			x					3 to 15	200°C operation	CELLMOS † CirTech	
CMOS, SOS	3	1	1	3			x					3 to 10		CELLMOS † CirTech	
	4	1	1	4			x					4 to 12		CELLMOS † CirTech	
	5	1	1	5			x					4 to 16		CELLMOS † CirTech	15
CMOS, Si-Gate	3	2				x		x	x			Avail. 1984		CMOS Mostek	
			1.8	x	x	x	x	x	x	x	3 to 10	Core μP, 1084		BHC Library Barvon (4626)	
Custom MOS Arrays CMOS, Metal-Gate	5	1			x	x	x		x	x	x	1.1 to 10	Low voltage and current	CMA 1000 Series CMA	
	CMOS, Si-Gate	3	1	1	3	x	x	x		x	x	2 to 7		CMOS Library CMA	
Fairchild Bipolar	3	2											Avail. 1H84	1 <sup>st</sup> L Fairchild	20
CMOS, Si-Gate	2	2	1										Avail. 1H84	CMOS Library Fairchild	
Fujitsu Microelectronics CMOS, Si-Gate	2.8	2	0	3.5					x	x		5		CMOS Library Fujitsu	
GTE Microcircuits CMOS, Si-Gate	3	2	2		x	x	x	x	x	x	x	5	Avail. 1084	ISO-CMOS Library GTEMicro	
	4	1	2	4.2					x	x	x	5		ISO-CMOS Library GTEMicro	
	5	1	2	5.2								5		ISO-CMOS Library GTEMicro	25

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

CUSTOM/SEMICUSTOM-Standard Cells (Cont'd)

Manufacturer Device Process Technology	Geometry ( $\mu$ m)	Interconnections		Delay 2-Input NAND Gate <sup>1</sup> (ns)	Cell Library Includes						Supply Voltage, V	Comments	Library or Process Name	Source	Line		
		Metal	Poly		A/D	D/A	Op Amp	Core $\mu$ P	RAM	ROM						PLA	
<b>Standard Cells</b>													<b>(Cont'd)</b>				
Harris Semiconductor CMOS, Si-Gate	3	1	1	3.8								1.5 to 7		SAJI-IV	† Harris	(4744)	
Hughes Aircraft CMOS, Si-Gate	2.5											4 to 10		HC MOS Library	† Hughes		
Integrated Circuit Systems CMOS, Metal-Gate	7.5	1		30						x		5 to 18		CMOS Library	† IntCirSys		
CMOS, Si-Gate	3	1	1	4						x		5		CMOS Library	† IntCirSys		
	5	1	1	10						x		5 to 10		CMOS Library	† IntCirSys		5
NMOS, Metal-Gate	4.5	1		20						x	x	5 to 10		NMOS Library	† IntCirSys		
	5	1	1	20						x	x	5 to 10		NMOS Library	† IntCirSys		
Intel CMOS, Si-Gate	2	1	1	1.0						x	x	5	Currently in Beta sites	CHMOS-1 Library	Intel		
Intersil CMOS, Si-Gate	3	2	1	<2								2 to 10	Avail. 1984	CMOS Library	Intersil		
Int'l Microcircuits CMOS, Si-Gate	3	2	1	2	x	x	x			x	x	3.5 to 8		CMOS Library	Int'l MicCir	(4747)	10
Int'l Microelectronic Prod CMOS, Si-Gate	3	2		2.5	x	x	x			x	x	5		Impel Library	† IMP	(4749)	
	5	1	1									3 to 12		Impel library	† IMP	(4749)	
	3		2		x	x	x			x	x	5		Impel library	† IMP	(4749)	
Master Logic CMOS, Metal-Gate	4	1		15	x	x	x					2 to 7		MALV Series	† MasterLogic		
CMOS, Si-Gate	4	1		3	x	x						2 to 7		MASG Series	† MasterLogic		15
MCE Semiconductor Bipolar, J. I.	5	1	1	12.5	x	x	x					to 75	Library includes some 75V devices	UniCELL	† MCE		
CMOS, Metal-Gate	5	1		10			x					1 to 18		UniCELL MGA	† MCE		
CMOS, Si-Gate	3.5	1	1	4.5			x					3 to 6		ULA Series	† MCE		
Microcircuit Technology NMOS, Si-Gate	3	1	1							x	x	5		NMOS Library	MicroTech		
CMOS, Si-Gate	3	1	1							x	x	3 to 7		CMOS Library	MicroTech		20
Mostek CMOS, Si-Gate	3	2					x			x	x		Avail. 1984	CMOS Library	Mostek		

<sup>1</sup> Delay conditions: 2-Input NAND driving 2 NANDs (FO = 2), 5V supply, 70°C, worst-case processing.

# IC MASTER

## CUSTOM/SEMICUSTOM-Standard Cells (Cont'd)

Manufacturer Device Process Technology	Geometry ( $\mu\text{m}$ )	Interconnections		Delay 2-Input NAND Gate* (ns)	Cell Library Includes						Supply Voltage, V	Comments	Library or Process Name	Source	Line	
		Metal	Poly		A/D	D/A	Op Amp	Core $\mu\text{P}$	RAM	ROM						PLA
<b>Standard Cells</b>															<b>(Cont'd)</b>	
NCM Corp CMOS, Metal-Gate	7.5	1	0	20			x		x	x	x	1.5 to 15		2000 Series	† NCM	
CMOS, Si-Gate	5	1	1	8.5			x		x	x	x	1.5 to 15		4000 Series	† NCM	
NCR Corp CMOS, Si-Gate	3	1	1	4					x	x	x	3 to 6		CMOS-2 Library	NCR	
NMOS, Si-Gate	4	1	1	6					x	x	x	5		VLSI-1	NCR	
Nitron CMOS, Metal-Gate	5	1		20	x	x	x				x	5 to 15		Nitrochip	† Nitron	5
CMOS, Si-Gate	5	1		5	x	x	x				x	3 to 10		Nitrochip	† Nitron	
Plessey Semiconductor CMOS, Si-Gate	2.5	2		1.4	x	x	x	x	x	x	x	3 to 7		CMOS	Plessey (4821)	
	5	1	1	6	x	x	x					3 to 7		CMOS	Plessey	
NMOS	5	1		20								5		CMOS	Plessey	
RCA Solid State CMOS, Si-Gate	3	1	1	5			x					3 to 7		PeCMOS-2	RCA (4824)	10
	5	1	1	10			x					3 to 7		PeCMOS-1	RCA (4824)	
	7	1	1	20			x					3 to 12		PeCMOS-C2L	RCA (4824)	
CMOS, SOS	3	1	1	4			x					3 to 12		SOS-2 Library	RCA (4824)	
	5	1	1	8			x					3 to 12		SOS-1 Library	RCA (4824)	
SGS-ATES Bipolar, MTL <sup>2</sup> V	6	2		25		x	x		x	x		2 to 10		Zodiac	SGS	15
Signetics Bipolar	5	2		3.5								5	Medium power	EPL-1	† Signetics (4855)	
		2		5								5	Low power	EPL-2	† Signetics (4855)	
		2		6								5		ISL	† Signetics (4855)	
CMOS, Si-Gate	3	2	1	8								2 to 6	High Speed	INTER	Signetics	
	4	1	1	8								3 to 15	Medium Speed	INTER	Signetics	20
Silicon Systems CMOS, Si-Gate	3	2	1	3	x	x	x	x	x	x	x	5		CMOS Library	SiliconSys	
Siliconix CMOS, Si-Gate	3	2	2	3		x	x		x	x		3 to 6	Avail. 4083	ISO-3 Library	Siliconix	
	5	2	2	5.5		x	x		x	x		3 to 14		ISO-5 Library	Siliconix	
Synertek CMOS, Si-Gate	3	1	1	8				x	x	x		3 to 6		NCMOS Library	† Synertek (4874)	
Texas Instruments CMOS, Si-Gate	3	1	1	4								5	Avail. 4083	CMOS Library	TI (4882)	25
VLSI Technology CMOS, Si-Gate	3	2	1	5				x	x	x	x	5	Configurable cells	CMOS Library	VTI (4931)	
NMOS	3	1	1	4				x	x	x	x	5	Configurable cells	NMOS Library	VTI (4931)	

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

CUSTOM/SEMICUSTOM-Standard Cells (Cont'd)

Manufacturer Device Process Technology	Geometry ( $\mu\text{m}$ )	Interconnections		Delay 2-Input NAND Gate' (ns)	Cell Library Includes							Supply Voltage, V	Comments	Library or Process Name	Source	Line	
		Metal	Poly		A/D	D/A	Op Amp	Core $\mu\text{P}$	RAM	ROM	PLA						
<b>Standard Cells</b>																	<b>(Cont'd)</b>
ZyMOS CMOS, Metal-Gate	6	1	1	12.2									1.1 to 6		Zy20000	† ZyMOS	
CMOS, Si-Gate	3	1	1	1.6					x	x	x	x	1.1 to 5.5		Zy50000	† ZyMOS	
	5	1	1	5.8	x	x	x	x	x	x	x	x	1.1 to 5.5		Zy40000	† ZyMOS	

\* Delay conditions: 2-Input NAND driving 2 NANDs (FO = 2), 5V supply, 70°C, worst-case processing.



CUSTOM/SEMICUSTOM-Programmable Logic

Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line	Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line
<b>Programmable Logic</b>							<b>Field Programmable Array Logic</b>						
Field Programmable Array Logic							30 nsF TS 5 (Cont'd)						
— TS 5							AMPAL16H8AM † AMD (4600) 60						
SN54PL20L8 † TI (4903,4914)							AMPAL16L8AM † AMD (4600)						
SN54PL20R4 † TI (4903,4914)							AMPAL16L8AM † AMD (4600)						
SN54PL20R6 † TI (4904,4914)							AMPAL16R4AM † AMD (4600)						
SN54PL20R8 † TI (4904,4914)							AMPAL16R4AM † AMD (4600)						
SN54PLR19L8 † TI (4906,4922) 5							AMPAL16R8AM † AMD (4600)						
SN54PLR19R4 † TI (4906,4922)							AMPAL16R8AM † AMD (4600)						
SN54PLR19R6 † TI (4907,4922)							PAL16L8AM † MMI (4768)						
SN54PLR19R8 † TI (4907,4922)							PAL16R4AM † MMI (4768)						
(353,4907,4922)							PAL16R6AM † MMI (4768)						
SN54PLT19L8 † TI							PAL16R8AM † MMI (4768)						
(353,4908,4926)							PAL20L8AM † MMI (4768) 70						
SN54PLT19R4 † TI							PAL20R4AM † MMI (4768)						
(353,4908,4926)							PAL20R6AM † MMI (4768)						
SN54PLT19R6 † TI 10							PAL20R8AM † MMI (4768)						
SN54PLT19R8 † TI							DMPAL10H8AM						
(353,4909,4926)							† National						
SN74PL20L8 TI (4903,4914)							DMPAL10L8AM						
SN74PL20R4 TI (4903,4914)							† National						
SN74PL20R6 TI (4904,4914) 15							DMPAL12H6AM						
SN74PL20R8 TI (4904,4914)							† National						
SN74PLR19L8 TI (4906,4922)							DMPAL12L6AM						
SN74PLR19R4 TI (4906,4922)							† National						
SN74PLR19R6 TI (4906,4922)							DMPAL14H4AM						
SN74PLR19R8 TI (4906,4922)							† National						
(353,4907,4922)							DMPAL14L4AM						
SN74PLR19R8 TI							† National						
(353,4907,4922) 20							DMPAL16C1AM						
SN74PLT19L8 TI							† National						
(353,4908,4926)							DMPAL16H2AM						
SN74PLT19R4 TI							† National						
(353,4908,4926)							DMPAL16L2AM						
SN74PLT19R6 TI							† National						
(353,4909,4926)							DMPAL16L8AM						
SN74PLT19R8 TI							† National						
(353,4909,4926)							DMPAL16R4AM						
25 nsF TS 5							† National						
AMPAL16H8AC AMD (4600) 25							DMPAL16R6AM						
AMPAL16H8AC AMD (4600)							† National						
AMPAL16L8AC AMD (4600)							DMPAL16R8AM						
AMPAL16L8AC AMD (4600)							† National						
AMPAL16R4AC AMD (4600)							DMPAL16R8AM						
AMPAL16R6AC AMD (4600)							† National						
AMPAL16R8AC AMD (4600) 30							SN54ALS16R4-1						
AMPAL22V10A AMD (4600)							† TI						
PAL16L8AC MMI (4768)							SN54PL16L8-1 † TI (4901,4910)						
PAL16R4AC MMI (4768)							SN54PL16R4-1 † TI (4901,4910)						
PAL16R6AC MMI (4768) 35							SN54PL16R6-1 † TI (4902,4910)						
PAL16R8AC MMI (4768)							SN54PL16R8-1 † TI (4902,4910)						
PAL20L8AC MMI (4768)							35 nsF TS 5						
PAL20R4AC MMI (4768)							AMPAL16H8C AMD (4600)						
PAL20R6AC MMI (4768)							AMPAL16H8C AMD (4600)						
PAL20R8AC MMI (4768) 40							AMPAL16H8C AMD (4600)						
DMPAL10H8AC National							AMPAL16H8C AMD (4600)						
DMPAL10L8AC National							AMPAL16L8C AMD (4600)						
DMPAL12H6AC National							AMPAL16L8C AMD (4600)						
DMPAL12L6AC National							AMPAL16L8C AMD (4600)						
DMPAL14H4AC National							AMPAL16L8C AMD (4600)						
DMPAL16C1AC National							AMPAL16L8C AMD (4600)						
DMPAL16H2AC National							AMPAL16R4C AMD (4600)						
DMPAL16L2AC National							AMPAL16R6C AMD (4600)						
DMPAL16L8AC National							AMPAL16R8C AMD (4600)						
DMPAL16R4AC National							AMPAL16R8C AMD (4600)						
DMPAL16R6AC National							AMPAL22V10 AMD (4600)						
DMPAL16R8AC National							AMPAL22V10 AMD (4600)						
DMPL14L4AC National							HPL77209-5 Harris (3958)						
SN74ALS16R4-1 TI							HPL77209/16L8-5 Harris (3958)						
SN74PL16L8-1 TI (4901,4910) 55							HPL77215/16R8-5						
SN74PL16R4-1 TI (4901,4910)							Harris (3958)						
SN74PL16R6-1 TI (4902,4910)							HPL77216/16P8-5 Harris (3958)						
SN74PL16R8-1 TI (4902,4910)							HPL77317/16L8-5						
AMPAL16H8AM † AMD (4600)							Harris (3959)						
(Continued)							HPL77318/16H8-5						
							Harris (3959)						
							HPL77319-5 Harris (3960)						
							(Continued)						

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

CUSTOM/SEMICUSTOM-Programmable Logic (Cont'd)

Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line	Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line
<b>Programmable Logic (Cont'd)</b>							<b>Field Programmable Array Logic (Cont'd)</b>						
	35 nsF	TS	5		(Cont'd)			40 nsF	TS	5			
				HPL77320-5	Harris	(3960)					DMPAL20C1M†	National	70
				PAL10H8C	MMI	(4768)					DMPAL20L2C	National	
				PAL10L8C	MMI	(4768)		45 nsF	TS	5	DMPAL20L2M†	National	
				PAL12H6C	MMI	(4768)	5				HPL77209-2	Harris	(3958)
				PAL12L6C	MMI	(4768)					HPL77209/16L8-2		
				PAL14H4C	MMI	(4768)						Harris	(3958)
				PAL14L4C	MMI	(4768)					HPL77215/16H8-2		
				PAL16H2C	MMI	(4768)						Harris	(3958)
				PAL16L2C	MMI	(4768)					HPL77216/16P8-2		
				PAL16L8A-2C	MMI	(4768)	10					Harris	(3958)
				PAL16L8C	MMI	(4768)					HPL77317-2	Harris	(3959)
				PAL16R4A-2C	MMI	(4768)					HPL77318-2	Harris	(3959)
				PAL16R4C	MMI	(4768)					HPL77319-2	Harris	(3960)
				PAL16R6A-2C	MMI	(4768)					PAL10H8M	MMI	(4768)
				PAL16R6C	MMI	(4768)	15				PAL10L8M	MMI	(4768)
				PAL16R8A-2C	MMI	(4768)					PAL12H6M	MMI	(4768)
				PAL16R8C	MMI	(4768)					PAL12L6M	MMI	(4768)
				DMPAL10H8C	National						PAL14H4M	MMI	(4768)
				DMPAL10L8C	National						PAL14L4M	MMI	(4768)
				DMPAL12H6C	National						PAL16A4M	MMI	(4768)
				DMPAL12L6C	National						PAL16C1M	MMI	(4768)
				DMPAL14H4C	National						PAL16H2M	MMI	(4768)
				DMPAL14L4C	National						PAL16L2M	MMI	(4768)
				DMPAL16C1C	National						PAL16L8M	MMI	(4768)
				DMPAL16H2C	National						PAL16R4M	MMI	(4768)
				DMPAL16L2C	National						PAL16R6M	MMI	(4768)
				DMPAL16L8C	National						PAL16R8M	MMI	(4768)
				DMPAL16R4C	National						PAL16X4M	MMI	(4768)
				DMPAL16R6C	National						DMPAL10H8M	National	
				DMPAL16R8C	National						DMPAL10L8M	National	95
	40 nsF	TS	5	AMPAL16H8LM†	AMD	(4600)					DMPAL12H6M	National	
				AMPAL16H8M†	AMD	(4600)					DMPAL12L6M	National	
				AMPAL16H8LM†	AMD	(4600)					DMPAL14H4M	National	
				AMPAL16H8M†	AMD	(4600)					DMPAL14L4M	National	
				AMPAL16L8LM†	AMD	(4600)	35				DMPAL16C1M	National	100
				AMPAL16L8M†	AMD	(4600)					DMPAL16H2M	National	
				AMPAL16L8LM†	AMD	(4600)					DMPAL16L2M	National	
				AMPAL16L8M†	AMD	(4600)					DMPAL16L8M	National	
				AMPAL16L8LM†	AMD	(4600)					DMPAL16R4M	National	105
				AMPAL16L8M†	AMD	(4600)					DMPAL16R6M	National	
				AMPAL16LDM†	AMD	(4600)					DMPAL16R8M	National	
				AMPAL16R4LM†	AMD	(4600)	40				PAL16L8-2C	MMI	(4768)
				AMPAL16R4M†	AMD	(4600)					PAL16L8A-2M†	MMI	(4768)
				AMPAL16R6LM†	AMD	(4600)					PAL16R4-2C	MMI	(4768)
				AMPAL16R6M†	AMD	(4600)					PAL16R4A-2M†	MMI	(4768)
				AMPAL16R8LM†	AMD	(4600)					PAL16R6-2C	MMI	(4768)
				AMPAL16R8M†	AMD	(4600)					PAL16R6A-2M†	MMI	(4768)
				PAL12L10C	MMI	(4768)	45				PAL16R8-2C	MMI	(4768)
				PAL12L10M†	MMI	(4768)					PAL16R8A-2M†	MMI	(4768)
				PAL14L8C	MMI	(4768)					PAL16R8A-2M†	MMI	(4768)
				PAL14L8M†	MMI	(4768)					PAL16R8A-2M†	MMI	(4768)
				PAL16A4C	MMI	(4768)					PAL16R8A-2M†	MMI	(4768)
				PAL16C1C	MMI	(4768)	50				PAL20L10M	MMI	(4768)
				PAL16L6C	MMI	(4768)					PAL20X10M	MMI	(4768)
				PAL16L6M†	MMI	(4768)					PAL20X4M	MMI	(4768)
				PAL16X4C	MMI	(4768)					PAL20X8M	MMI	(4768)
				PAL18L4C	MMI	(4768)					DMPAL209X4C	National	
				PAL18L4M†	MMI	(4768)					DMPAL20L10C	National	120
				PAL20C1C	MMI	(4768)	55				DMPAL20X10C	National	
				PAL20C1M†	MMI	(4768)					DMPAL20X8C	National	
				PAL20L2C	MMI	(4768)					PAL16L8A-4C	MMI	(4768)
				PAL20L2M†	MMI	(4768)					PAL16R4A-4C	MMI	(4768)
				DMPAL12L10C	National						PAL16R6A-4C	MMI	(4768)
				DMPAL12L10M†	National						PAL16R8A-4C	MMI	(4768)
				DMPAL14L8C	National						PAL10H3-2C	MMI	(4768)
				DMPAL14L8M†	National						PAL10L8-2C	MMI	(4768)
				DMPAL16L6C	National						PAL12H6-2C	MMI	(4768)
				DMPAL16L6M†	National						PAL12L6-2C	MMI	(4768)
				DMPAL18L4C	National						PAL14H4-2C	MMI	(4768)
				DMPAL18L4M†	National						PAL14L4-2C	MMI	(4768)
				DMPAL20C1C	National						PAL16C1-2C	MMI	(4768)
											PAL16H2-2C	MMI	(4768)

† Military Temperature Range (-55° to 125°C)  
OC—Open Collector

ns\*—Nanoseconds Typical

nsF—Nanoseconds over Full Temperature Range  
TS—Three-State

nsR—Nanoseconds at Room Temperature  
OE—Open Emitter

CUSTOM/SEMICUSTOM-Programmable Logic (Cont'd)

Organi- zation	Propa- gation Time	Output	Supply Voltage, V	Device	Source	Line	Organi- zation	Propa- gation Time	Output	Supply Voltage, V	Device	Source	Line
<b>Programmable Logic (Cont'd)</b>							<b>Field Programmable Identity Comparator, 8-Bit and 4-Bit Comparator</b>						
Field Programmable Array Logic							5						
60 nsF	TS		5							5	SN54ALS527 † TI	(1045)	55
											SN74ALS527 TI	(1045)	
Field Programmable Logic Array, 14 Inputs, 32 Product Terms, Six Sum Terms							10 ns * OC 5						
										5	SN54PL840 † TI	(4905,4918)	
											SN74PL840 TI	(4905,4918)	
										5	SN54PL839 † TI	(4905,4918)	60
											SN74PL839 TI	(4905,4918)	
Field Programmable Logic Array, 16 Inputs, 48 Product Terms, 8 Outputs							50 nsF OC 5						
65 nsF	TS		5							5	MC82S101C Motorola	(4153)	
											N82S101 Signetics		
										5	MC82S100C Motorola	(4153)	
											N82S100 Signetics		
										5	MC82S101M † Motorola	(532,4153)	65
											S82S101 † Signetics		
										5	MC82S100M † Motorola	(532,4153)	
											S82S100 † Signetics		
Field Programmable Logic Array, 24 Inputs, 72 Product Terms, 16 Outputs							200 nsF TS 5						
											μPB450	NEC	
Field Programmable Logic Array, 48 Product Terms, 14 Inputs							100 nsF OC 5						
											IM5200	Intersil	70
Field Programmable Logic Array, 8 Inputs, 32 AND Gates, 10 OR Gates, 10 I/O Lines							30 nsF OC 5						
											N82S152A Signetics		
										5	N82S153A Signetics	(4146,4148)	
										5	N82S152 Signetics		
										5	HPL77153-5 Harris	(3957)	
										5	N82S153 Signetics	(4146)	75
										5	S82S152A † Signetics		
										5	S82S153A † Signetics	(532,4146,4148)	
										5	S82S152 † Signetics		
										5	HPL77153-2 † Harris	(3957)	
										5	S82S153 † Signetics	(532,4146)	80
Field Programmable Logic Element (PROM), 1024x4							35 nsF TS 5						
											63S441A MMI	(4026)	
										5	53S441A † MMI	(4026)	
Field Programmable Logic Element (PROM), 2048x4							35 nsF TS 5						
											63S841A MMI	(4026)	
										5	53S841A † MMI	(4026)	
Field Programmable Logic Element (PROM), 32x8							25 nsF TS 5						
											63S080 MMI	(4026)	
											63S081 MMI	(4026)	85
										5	53S080 † MMI	(4026)	
										5	53S081 † MMI	(4026)	
Field Programmable Logic Element (PROM), 4096x4							35 nsF TS 5						
											63S1641A MMI	(4026)	
										5	53S1641A † MMI	(4026)	90
Field Programmable Logic Sequencer, 12 Inputs, 32 Product Terms, 6 Outputs							45 nsF * OC 5						
											SN54PL335 † TI		
											SN74PL335 TI		
										5	SN54PL333 † TI		
											SN74PL333 TI		
Field Programmable Logic Sequencer, 12 Inputs, 50 Product Terms, 6 Outputs							35 ns * TS 5						
											SN74S331 TI		95
										5	SN74S330 TI		

† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

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CUSTOM/SEMICUSTOM

CUSTOM/SEMICUSTOM—Programmable Logic (Cont'd)

Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line	Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line	
<b>Programmable Logic</b>							<b>Hard Array Logic</b>							
<b>(Cont'd)</b>							<b>(Cont'd)</b>							
Field Programmable Logic Sequencer, 16 Inputs, 48 Transition Terms, 8 Outputs							35 nsF TS 5							
60 nsF	OC	5	N82S104A	Signetics							SP810H8	SPI	(4826)	
			S82S104A	† Signetics	(532)						SP810L8	SPI	(4826)	
	TS	5	N82S105A	Signetics	(4157)						SP812H6	SPI	(4826)	
			S82S105A	† Signetics	(532,4157)						SP812L6	SPI	(4826)	
90 nsF	OC	5	N82S104	Signetics		5					SP814H4	SPI	(4826)	
			S82S104	† Signetics	(532)						SP814L4	SPI	(4826)	
	TS	5	N82S105	Signetics							SP816H2	SPI	(4826)	
			S82S105	† Signetics	(532)						SP816L2	SPI	(4826)	
Field Programmable Logic Sequencer, 4 Inputs, 32 AND Gates, 21 OR Gates, 4 I/O Lines							FCT20x4						SI-Fab	(4827)
35 ns *	OC	5	S82S158	† Signetics	(532)						HCT10H8	SI-Fab	(4827)	
	TS	5	S82S159	† Signetics	(532,4148)						HCT10L8	SI-Fab	(4827)	
40 nsF	OC	5	N82S158	Signetics		10					HCT12L10	SI-Fab	(4827)	
			S82S159	Signetics	(4148)						HCT12L8	SI-Fab	(4827)	
	TS	5	N82S158	Signetics							HCT14H4	SI-Fab	(4827)	
			N82S159	Signetics							HCT14L4	SI-Fab	(4827)	
Field Programmable Logic Sequencer, 4 Inputs, 32 AND Gates, 21 OR Gates, 6 I/O Lines							HCT14L8						SI-Fab	(4827)
35 ns *	OC	5	S82S156	† Signetics	(532)						HCT16A4	SI-Fab	(4827)	
	TS	5	S82S157	† Signetics	(532)						HCT16C1	SI-Fab	(4827)	
40 nsF	OC	5	N82S156	Signetics		15					HCT16H2	SI-Fab	(4827)	
			N82S157	Signetics							HCT16L2	SI-Fab	(4827)	
	TS	5	N82S156	Signetics							HCT16L6	SI-Fab	(4827)	
			N82S157	Signetics							HCT16L8	SI-Fab	(4827)	
Field Programmable Logic Sequencer, 4 Inputs, 32 AND Gates, 21 OR Gates, 8 I/O Lines							HCT16R4						SI-Fab	(4827)
35 ns *	OC	5	S82S154	† Signetics	(532)						HCT16R8	SI-Fab	(4827)	
	TS	5	S82S155	† Signetics	(532)						HCT16R8	SI-Fab	(4827)	
40 nsF	OC	5	N82S154	Signetics		20					HCT16X4	SI-Fab	(4827)	
			N82S155	Signetics							HCT18L4	SI-Fab	(4827)	
	TS	5	N82S154	Signetics							HCT20C1	SI-Fab	(4827)	
			N82S155	Signetics							HCT20L10	SI-Fab	(4827)	
Field Programmable Multiplexer, (10 inputs, four 10x8 programmable arrays, four 8:1 addressable multiplexers, four outputs)							HCT20L2						SI-Fab	(4827)
55 nsF	TS	5	29693C	Raytheon							HCT20x10	SI-Fab	(4827)	
75 nsF	TS	5	29693M	† Raytheon							HCT20x8	SI-Fab	(4827)	
Hard Array Logic							ZX-CAL10H8						Zytrex	
25 nsF	TS	5	HAL16L8AC	MMI	(4776,4784,4786)						ZX-CAL10L8	Zytrex		
			HAL16R4AC	MMI	(4776,4784,4786)						ZX-CAL12H6	Zytrex		
			HAL16R6AC	MMI	(4776,4784,4786)	25					ZX-CAL12L6	Zytrex		
			HAL16R8AC	MMI	(4776,4784,4786)						ZX-CAL14H4	Zytrex		
			HAL20L8AC	MMI	(4776,4787)						ZX-CAL14L4	Zytrex		
			HAL20R4AC	MMI	(4776,4787)						ZX-CAL16H2	Zytrex		
			HAL20R6AC	MMI	(4776,4787)						ZX-CAL16L2	Zytrex		
			HAL20R8AC	MMI	(4776,4787)	30					ZX-CAL16L8	Zytrex		
30 nsF	TS	5	HAL20L8AM	† MMI	(4776)						ZX-CAL16R4	Zytrex		
			HAL20R4AM	† MMI	(4776)						ZX-CAL16R6	Zytrex		
			HAL20R6AM	† MMI	(4776)						ZX-CAL16R8	Zytrex		
			HAL20R8AM	† MMI	(4776)						ZX-CAL16R8	Zytrex		
35 nsF	TS	5	C-HAL16L8	MMI		35					HAL12L10C	MMI	(4776,4783)	
			C-HAL16R4	MMI							HAL14L8C	MMI	(4776,4783)	
			C-HAL16R6	MMI							HAL16A4C	MMI	(4776,4784)	
			C-HAL16R8	MMI							HAL16C1C	MMI	(4776,4782)	
			HAL10H8C	MMI	(4776,4782)						HAL16L6C	MMI	(4776,4783)	
			HAL10L8C	MMI	(4776,4782)						HAL16X4C	MMI	(4776,4784)	
			HAL12H6C	MMI	(4776,4782)						HAL18L4C	MMI	(4776,4783)	
			HAL12L6C	MMI	(4776,4782)						HAL20C1C	MMI	(4776,4783)	
			HAL14H4C	MMI	(4776,4782)						HAL20L2C	MMI	(4776,4783)	
			HAL14L4C	MMI	(4776,4782)						SP816C1	SPI	(4826)	
			HAL16H2C	MMI	(4776,4782)						SP810H8 † MMI (4776,4782)			
			HAL16L2C	MMI	(4776,4782)						HAL10L8M † MMI (4776,4782)			
			HAL16L8C	MMI	(4776,4784)						HAL12H6M † MMI (4776,4782)			
			HAL16R4C	MMI	(4776,4784)						HAL12L10M † MMI (4776,4783)			
			HAL16R6C	MMI	(4776,4784)						HAL12L6M † MMI (4776,4782)			
			HAL16R8C	MMI	(4776,4784)						HAL14H4M † MMI (4776,4782)			
											HAL14L4M † MMI (4776,4782)			
											HAL16A4M † MMI (4776,4784)			
											HAL16C1M † MMI (4776,4782)			
											HAL16H2M † MMI (4776,4782)			
											HAL16L2M † MMI (4776,4782)			

† Military Temperature Range (-55° to 125°C) ns\*—Nanoseconds Typical nsF—Nanoseconds over Full Temperature Range nsR—Nanoseconds at Room Temperature OC—Open Collector TS—Three-State OE—Open Emitter

# IC MASTER

## CUSTOM/SEMICUSTOM-Programmable Logic (Cont'd)

Organization	Propagation Time	Output	Supply Voltage, V	Device	Source	Line
<b>Programmable Logic</b>				<b>(Cont'd)</b>		
Hard Array Logic				(Cont'd)		
45 nsF	TS		5	HAL16L6M † MMI (4776.4783)		5
				HAL16L8M † MMI (4776.4784)		
				HAL16R4M † MMI (4776.4784)		
				HAL16R6M † MMI (4776.4784)		
				HAL16R8M † MMI (4776.4784)		
				HAL16X4M † MMI (4776.4784)		
				HAL18L4M † MMI (4776.4783)		
				HAL20L2M † MMI (4776.4783)		
				HAL24L8M † MMI		
50 nsF	TS		5	HAL16L8-2C MMI (4776.4784.4789)		10
				HAL16R4-2C MMI (4776.4784.4789)		
				HAL16R6-2C MMI (4776.4784.4789)		
				HAL16R8-2C MMI (4776.4784.4789)		
				HAL20L10C MMI (4776.4785)		
				HAL20X10C MMI (4776.4785)		
				HAL20X4C MMI (4776.4785)		
				HAL20X8C MMI (4776.4785)		
60 nsF	TS		5	HAL10H8-2C MMI (4776.4782.4788)		20
				HAL10L8-2C MMI (4776.4782.4788)		
				HAL12H6-2C MMI (4776.4782.4788)		
				HAL12L6-2C MMI (4776.4782.4788)		
				HAL14H4-2C MMI (4776.4782.4788)		
				HAL14L4-2C MMI (4776.4782.4788)		
				HAL16C1-2C MMI (4776.4782.4788)		
				HAL16H2-2C MMI (4776.4782)		
				HAL16L2-2C MMI (4776.4782.4788)		
				HAL20C1M † MMI (4776.4783)		25
				HAL20L10M † MMI (4776.4785)		
				HAL20X10M † MMI (4776.4785)		
				HAL20X4M † MMI (4776.4785)		
				HAL20X8M † MMI (4776.4785)		
75 nsF	TS		5	HAL16L8-4C MMI (4776.4784.4790)		30
				HAL16R4-4C MMI (4776.4784.4790)		
				HAL16R6-4C MMI (4776.4784.4790)		
				HAL16R8-4C MMI (4776.4784.4790)		
Programmed at Factory, 14 Inputs, 96 Product Terms						35
150 nsF	—		5	DM7575 † National		
				DM7576 † National		
				DM8575 National		
				DM8576 National		

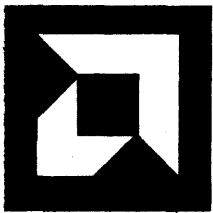
† Military Temperature Range (-55° to 125°C)

\* Typical Value

Bold face indicates additional data is provided on the page noted.

# ABBREVIATIONS OF COMPANY NAMES

<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.			<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Harris</b>	Harris Semiconductor	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Heurikon</b>	Heurikon Corp.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Hitachi</b>	Hitachi America, Ltd.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>Holt</b>	Holt Inc.		
<b>Analogic</b>	Analogic	<b>HP</b>	Hewlett-Packard		
<b>Analog Sys</b>	Analog Systems	<b>Hughes</b>	Hughes Aircraft, Solid State Products		
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Panasonic</b>	Panasonic
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>Pico Design</b>	Pico Design
<b>APM</b>	Applied Microsystems Corp.			<b>Polycore</b>	Polycore Electronics
<b>Appl Sys</b>	Applied Systems Corp.			<b>Plessey</b>	Plessey Semiconductors
<b>APT</b>	Applied Microtechnology			<b>PMI</b>	Precision Monolithics, Inc.
<b>Aptek</b>	Aptek Microsystems			<b>PragDes</b>	Pragmatic Design Inc.
<b>Array Tech</b>	Array Technology			<b>Pro-Log</b>	Pro-Log Corp.
<b>AWI</b>	AWI Electronics				
				<b>Quay</b>	Quay Corp.
<b>Barvon</b>	Barvon Research	<b>ICC</b>	International Cybernetics	<b>Raytheon</b>	Raytheon Semiconductor
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>IDT</b>	Integrated Device Technology	<b>RCA</b>	RCA Solid State Division
<b>Burr-Brown</b>	Burr-Brown	<b>IMI</b>	International Microcircuits, Inc.	<b>RCI Data</b>	RCI Data
		<b>IMP</b>	International Microelectronic Products	<b>RELMS</b>	Relational Memory Systems
<b>CAE</b>	Computer Aided Engineering	<b>IMS</b>	Industrial MicroSystems Inc.	<b>Reticon</b>	Reticon
<b>Cal Devices</b>	California Devices	<b>Infosphere</b>	Infosphere	<b>RIFA</b>	RIFA
<b>Cermetek</b>	Cermetek	<b>Inmos</b>	Inmos	<b>Rockwell</b>	Rockwell, Microelectronic Devices
<b>CGRS</b>	CGRS Microtech Inc.	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>RTC</b>	Riehl Time Corporation
<b>Cherry</b>	Cherry Semiconductor	<b>IntCirSys</b>	Integrated Circuit Systems		
<b>CIC</b>	Custom Integrated Circuits	<b>IntCompSys</b>	Integrated Computer Systems		
<b>CirTech</b>	Circuit Technology	<b>Int Tech</b>	Integrated Technology Corp.		
<b>Citel</b>	Citel	<b>Intech</b>	Intech Microcircuits Div.		
<b>Comlinear</b>	Comlinear Corporation	<b>Intel</b>	Intel		
<b>CMA</b>	Custom MOS Arrays	<b>Interdesign</b>	Interdesign		
<b>Comark</b>	Comark Corp.	<b>Intersil</b>	Intersil		
<b>Comdial</b>	Comdial Semiconductor	<b>Intronics</b>	Intronics		
<b>Comp Auto</b>	Computer Automation	<b>ITT</b>	ITT Semiconductors		
<b>Compas</b>	Compas Microsystems			<b>Sanyo</b>	Sanyo
<b>Cont Logic</b>	Control Logic Inc.	<b>Kinetic Sys</b>	Kinetic Systems	<b>SBE</b>	SBE, Inc.
<b>Control Sys</b>	Control Systems Microsystems Div.	<b>Kontron</b>	Kontron Electronics	<b>SEEQ</b>	SEEQ Technology, Inc.
<b>CreMicro</b>	Creative Micro Systems			<b>SPI</b>	Semi Processes Inc.
<b>Cromemco</b>	Cromemco, Inc.	<b>Lambda</b>	Lambda Semiconductor	<b>Siemens</b>	Siemens
<b>Cubit</b>	Cubit Inc.	<b>Linear Tech</b>	Linear Technology	<b>Si-Fab</b>	Si-Fab
<b>Curtis</b>	Curtis Electro Devices, Inc.	<b>LSI Comp</b>	LSI Computer Systems	<b>Signetics</b>	Signetics
<b>Cybernetic</b>	Cybernetic Micro Systems	<b>LSI Logic</b>	LSI Logic Corporation	<b>SGS</b>	SGS Semiconductor
<b>Cybersys</b>	Cybersystems			<b>Sharp</b>	Sharp
<b>Cybertek</b>	Cybertek Inc.	<b>Master Logic</b>	Master Logic Corporation	<b>Silicon G</b>	Silicon General
		<b>Matrix</b>	Matrix Corp.	<b>Siliconix</b>	Siliconix
		<b>Matrox</b>	Matrox Electronic Systems	<b>Silicon Sys</b>	Silicon Systems Inc.
		<b>MCC</b>	Micro-Computer Control	<b>Siltronics</b>	Siltronics
		<b>MCE</b>	MCE Electronics	<b>SMC</b>	Standard Microsystems Corp.
		<b>Micrel</b>	Micrel	<b>S MOS</b>	S MOS Systems
		<b>Micro Innov</b>	Micro Innovators	<b>Solarise</b>	Solarise Enterprises
		<b>Micropac</b>	Micropac Industries	<b>Solitron</b>	Solitron Devices
		<b>Micro Net</b>	Micro Networks	<b>Sprague</b>	Sprague Electric Company
		<b>Micro Pwr</b>	Micro Power Systems	<b>SSM</b>	Spid State Micro Technology for Music
		<b>Micro Sci</b>	Micro Sciences Corp.	<b>SSS</b>	Solid State Scientific
		<b>Micro Tech</b>	Microcircuits Technology	<b>Stag</b>	Stag Microsystems
		<b>Micro-Link</b>	Micro-Link Corporation	<b>STC</b>	Storage Technology Corp.
		<b>Micron</b>	Micron Technology	<b>STD</b>	STD Microsystems
		<b>MilerTron</b>	MilerTronics	<b>Struc Des</b>	Structured Design Inc.
		<b>Miller</b>	Miller Technology	<b>Stynetic</b>	Stynetic Systems
		<b>Mitel</b>	Mitel Semiconductor	<b>Sunrise</b>	Sunrise Electronics
		<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Sunshine</b>	Sunshine Semiconductor
		<b>MMI</b>	Monolithic Memories, Inc.	<b>Supertex</b>	Supertex Inc.
				<b>Symtek</b>	Symtek Corp.
		<b>Mostek</b>	Monolithic Systems Corp.	<b>Synertek</b>	Synertek
		<b>Motorola</b>	Motorola Semiconductor	<b>Sys Innov</b>	Systems Innovations
		<b>MRC</b>	MRC Systems		
		<b>Murray</b>	Murray Consulting	<b>Tau Zero</b>	Tau Zero Inc.
		<b>Monosil</b>		<b>Technitrol</b>	Technitrol
				<b>Tektronix</b>	Tektronix
		<b>National</b>	National Semiconductor	<b>Teledyne C</b>	Teledyne Crystalonics
		<b>NCM</b>	NCM Corp.	<b>Teledyne P</b>	Teledyne Philbrick
		<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Teledyne S</b>	Teledyne Semiconductor
				<b>Telefunken</b>	Telefunken
		<b>NEC</b>	NEC Electronics	<b>Telmos</b>	Telmos
		<b>Nitron</b>	Nitron	<b>Teltone</b>	Teltone Corporation
				<b>TI</b>	Texas Instruments
				<b>Third Domain</b>	Third Domain
				<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
				<b>Toshiba</b>	Toshiba America
				<b>Trans-Data</b>	Trans-Data
				<b>TRW</b>	TRW LSI Products
				<b>Unitrode</b>	Unitrode
				<b>Universal</b>	Universal Semiconductor, Inc.
				<b>Varix</b>	Varix Corp.
				<b>VLSI Design</b>	VLSI Design Associates
				<b>VTI</b>	VLSI Technology, Inc.
				<b>Votrax</b>	Votrax
				<b>Weitek</b>	Weitek Corporation
				<b>Western</b>	Western Digital
				<b>Wintek</b>	Wintek Corp.
				<b>Xicor</b>	Xicor, Inc.
				<b>Xycom</b>	Xycom
				<b>Zendex</b>	Zendex Corp.
				<b>Zilog</b>	Zilog
				<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrex</b>	Zytrex Corp.
<b>Fairchild</b>	Fairchild				
<b>Ferranti</b>	Ferranti Electric				
<b>Force</b>	Force Computers				
<b>Fujitsu A</b>	Fujitsu America				
<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.				



# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements

### Features of PALs

- User customizable, high performance logic building blocks
- Custom logic patterns may be generated in minutes with PROM type programmers
- Easy to use software design aids available (PALASM\*)
- Improves performance and reduces board area and cost of existing TTL SSI/MSI designs
- Aids creation of new system architectures through interactive design techniques
- Security fuse prevents copying of logic by competitors
- Slim 20 and 24-pin DIP packages

### Advantages of AMD PALs

- IMOXTM oxide isolated technology insures industry's fastest (12ns typ) "A" versions and fastest half-power (24ns typ) "L" versions
- Platinum-silicide fuses and added test words insure programming yields > 98%
- Functional yield after programming > 99%
- Reliability assured through more than 40 billion fuse hours of life testing with no failures
- Full AC and DC parameter testing at the factory through on-board testing circuitry
- Power-up reset simplifies state machine design
- Industry leading quality guarantees

### AMD PAL Speed/Power Families

Family	t <sub>pd</sub> ns (Max)	t <sub>g</sub> <sup>(1)</sup> ns (Max)	t <sub>co</sub> <sup>(1)</sup> ns (Max)	I <sub>cc</sub> <sup>(2)</sup> mA (Max)	I <sub>OL</sub> mA (Min)
High Speed, "A"	25	20	15	155	24
Standard	35	30	25	155	24
Half Power, "L"	35	30	25	80	24

(1) Sequential functions.

(2) Combinatorial functions.

### AMD PAL FUNCTIONS

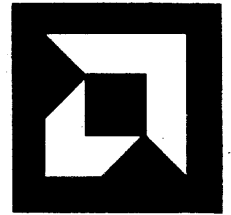
Part Number	Array Inputs	Logic	OE	Outputs	Package Pins
16R8	Eight Dedicated Eight Feedback	Eight 8-Wide AND-OR	Dedicated	Registered Inverting	20
16R6	Eight Dedicated Six Feedback Two Bidirectional	Six 8-Wide AND-OR	Dedicated	Registered Inverting	20
		Two 7-Wide AND-OR-INVERT	Programmable	Bidirectional	
16R4	Eight Dedicated Four Feedback Four Bidirectional	Four 8-Wide AND-OR	Dedicated	Registered Inverting	20
		Four 7-Wide AND-OR-INVERT	Programmable	Bidirectional	
16L8	Ten Dedicated Six Bidirectional	Eight 7-Wide AND-OR-INVERT	Programmable	Six Bidirectional Two Dedicated	20
16H8	Ten Dedicated Six Bidirectional	Eight 7-Wide AND-OR	Programmable	Six Bidirectional Two Dedicated	20
16LD8	Ten Dedicated Six Bidirectional	Eight 8-Wide AND-OR-INVERT	-	Dedicated	20
16HD8	Ten Dedicated Six Bidirectional	Eight 8-Wide AND-OR	-	Dedicated	20
22V10	Twelve Dedicated Ten Bidirectional/ Feedback	Ten 12 (Average)-Wide AND-OR/ AND-OR-INVERT	Programmable	Ten Bidirectional/Registered Programmable Polarity	24

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\*PAL is a registered trademark of Monolithic Memories, Inc.

\*PALASM is a registered trademark of Monolithic Memories, Inc.

# Advanced Micro Devices



## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements

### MAXIMUM RATINGS (Above which the useful life may be impaired)

Storage Temperature	-65 to +150°C
Temperature (Ambient) Under Bias	-55 to +125°C
Supply Voltage to Ground Potential (Pin 20 to Pin 10) Continuous	-0.5 to +7V
DC Voltage Applied to Outputs (Except During Programming)	-0.5V to +V <sub>CC</sub> max
DC Voltage Applied to Outputs During Programming	21V
Output Current Into Outputs During Programming (Max Duration of 1 sec)	200mA
DC Input Voltage	-0.5 to +5.5V
DC Input Current	-30 to +5mA

### OPERATING RANGE

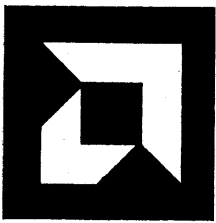
Parameters	Description	Commercial		Military		Units
		Min	Max	Min	Max	
V <sub>CC</sub>	Supply Voltage	4.75	5.25	4.50	5.50	V
T <sub>A</sub>	Operating Free Air Temperature	0	75	-55		°C
T <sub>C</sub>	Operating Case Temperature				125	°C

### ELECTRICAL CHARACTERISTICS OVER OPERATING RANGE (Unless Otherwise Noted)

Parameters	Description	Test Conditions	Typ			Units
			Min	(Note 1)	Max	
V <sub>OH</sub>	Output HIGH Voltage	V <sub>CC</sub> = MIN, V <sub>IN</sub> = V <sub>IH</sub> or V <sub>IL</sub> I <sub>OH</sub> = -3.2mA COM'L I <sub>OH</sub> = -2mA MIL	2.4	3.5		Volts
V <sub>OL</sub>	Output LOW Voltage	V <sub>CC</sub> = MIN, V <sub>IN</sub> = V <sub>IH</sub> or V <sub>IL</sub> I <sub>OL</sub> = 24mA COM'L I <sub>OL</sub> = 12mA MIL			0.50	Volts
V <sub>IH</sub> (Note 2)	Input HIGH Level	Guaranteed input logical HIGH voltage for all inputs	2.0			Volts
V <sub>IL</sub> (Note 2)	Input LOW Level	Guaranteed input logical LOW voltage for all inputs			0.8	Volts
I <sub>IL</sub>	Input LOW Current	V <sub>CC</sub> = MAX, V <sub>IN</sub> = 0.40V		-20	-250	μA
I <sub>IH</sub>	Input HIGH Current	V <sub>CC</sub> = MAX, V <sub>IN</sub> = 2.7V			25	μA
I <sub>I</sub>	Input HIGH Current	V <sub>CC</sub> = MAX, V <sub>IN</sub> = 5.5V			1.0	mA
I <sub>SC</sub>	Output Short Circuit Current	V <sub>CC</sub> = MAX, V <sub>OUT</sub> = 0.5V (Note 3)	-30	-60	-90	mA
I <sub>CC</sub>	Power Supply Current	All inputs = GND, V <sub>CC</sub> = MAX 16L8, 16H8, 16HD8, 16LD8 16L8A, 16H8A, 16HD8A, 16LD8A 16R8, 16R6, 16R4 16R8A, 16R6A, 16R4A		110	155	mA
V <sub>I</sub>	Input Clamp Voltage	V <sub>CC</sub> = MIN, I <sub>IN</sub> = -18mA		-0.9	-1.2	Volts
I <sub>OZH</sub>	Output Leakage Current	V <sub>CC</sub> = MAX, V <sub>IL</sub> = 0.8V V <sub>O</sub> = 2.7V			100	μA
I <sub>OZL</sub>	Output Leakage Current (Note 4)	V <sub>CC</sub> = MAX, V <sub>IL</sub> = 0.8V V <sub>O</sub> = 0.4V			-100	μA
C <sub>IN</sub>	Input Capacitance	V <sub>IN</sub> = 2.0V @ f = 1MHz (Note 5)		6		pF
C <sub>OUT</sub>	Output Capacitance	V <sub>OUT</sub> = 2.0V @ f = 1MHz (Note 5)		9		pF

- Notes: 1. Typical limits are at V<sub>CC</sub> = 5.0V and T<sub>A</sub> = 25°C.  
 2. These are absolute values with respect to device ground and all overshoots due to system or tester noise are included.  
 3. Not more than one output should be tested at a time. Duration of the short circuit should not be more than one second. V<sub>OUT</sub> = 0.5V has been chosen to avoid test problems caused by tester ground degradation.  
 4. I/O pin leakage is the worst case of I<sub>OZX</sub> or I<sub>I<sub>X</sub></sub> (where X = H or L).  
 5. These parameters are not 100% tested, but are periodically sampled.





# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements

### SWITCHING CHARACTERISTICS OVER OPERATING RANGE (Unless otherwise noted) HIGH SPEED

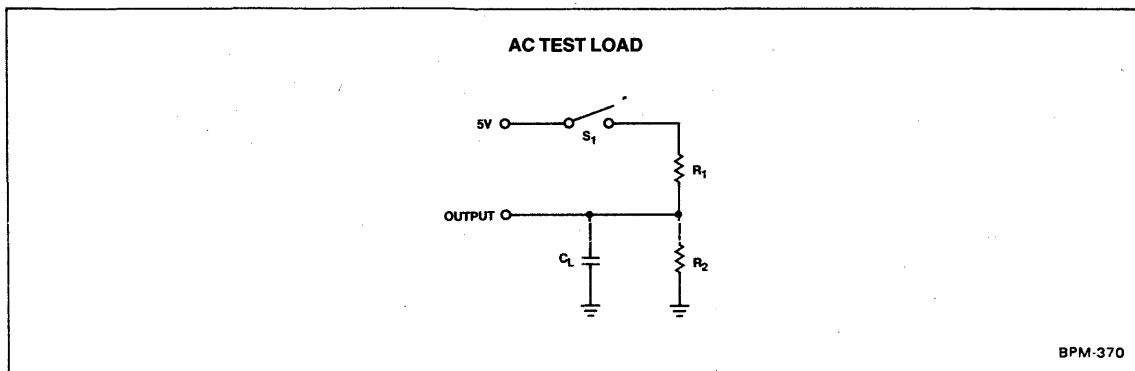
Parameters	Description	Test Conditions	Typ (Note 1)	COM'L		MIL		Units
				Min	Max	Min	Max	
t <sub>PD</sub>	Input or Feedback to Non-Registered Output 16L8A, 16R6A, 16R4A, 16LD8A, 16H8A, 16HD8A	COM'L R <sub>1</sub> = 200 R <sub>2</sub> = 390	12		25		30	ns
t <sub>EA</sub>	Input to Output Enable 16L8A, 16R6A, 16R4A, 16H8A		12		25		30	ns
t <sub>ER</sub>	Input to Output Disable 16L8A, 16R6A, 16R4A, 16H8A		12		25		30	ns
t <sub>PZX</sub>	Pin 11 to Output Enable 16R8A, 16R6A, 16R4A		8		20		25	ns
t <sub>PXZ</sub>	Pin 11 to Output Disable 16R8A, 16R6A, 16R4A		8		20		25	ns
t <sub>CO</sub>	Clock to Output 16R8A, 16R6A, 16R4A		8		15		20	ns
t <sub>s</sub>	Input or Feedback Setup Time 16R8A, 16R6A, 16R4A		MIL R <sub>1</sub> = 390 R <sub>2</sub> = 750	10	20		25	ns
t <sub>H</sub>	Hold Time 16R8A, 16R6A, 16R4A			-10	0		0	ns
t <sub>p</sub>	Clock Period				35		45	ns
t <sub>w</sub>	Clock Width				15		20	ns
f <sub>MAX</sub>	Maximum Frequency				28.5		22	MHz

- Notes: 1. Typical limits are at V<sub>CC</sub> = 5.0V and T<sub>A</sub> = 25°C.  
 2. t<sub>PD</sub> is tested with switch S<sub>1</sub> closed and C<sub>L</sub> = 50pF.  
 3. For three-state outputs, output enable times are tested with C<sub>L</sub> = 50pF to the 1.5V level; S<sub>1</sub> is open for high impedance to HIGH tests and closed for high impedance to LOW tests. Output disable times are tested with C<sub>L</sub> = 5pF. HIGH to high impedance tests are made to an output voltage of V<sub>OH</sub> - 0.5V with S<sub>1</sub> open; LOW to high impedance tests are made to the V<sub>OL</sub> + 0.5V level with S<sub>1</sub> closed.

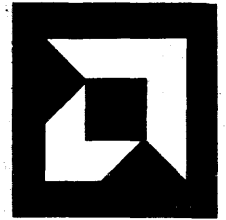
### SWITCHING CHARACTERISTICS OVER OPERATING RANGE (Unless otherwise noted) STANDARD SPEED

Parameters	Description	Test Conditions	Typ (Note 1)	COM'L		MIL		Units
				Min	Max	Min	Max	
t <sub>PD</sub>	Input or Feedback to Non-Registered Output 16L8, 16R6, 16R4, 16LD8, 16H8, 16HD8	COM'L R <sub>1</sub> = 200 R <sub>2</sub> = 390	17		35		40	ns
t <sub>EA</sub>	Input to Output Enable 16L8, 16R6, 16R4, 16H8		17		35		40	ns
t <sub>ER</sub>	Input to Output Disable 16L8, 16R6, 16R4, 16H8		17		35		40	ns
t <sub>PZX</sub>	Pin 11 to Output Enable 16R8, 16R6, 16R4		12		25		25	ns
t <sub>PXZ</sub>	Pin 11 to Output Disable 16R8, 16R6, 16R4		12		25		25	ns
t <sub>CO</sub>	Clock to Output 16R8, 16R6, 16R4		12		25		25	ns
t <sub>s</sub>	Input or Feedback Setup Time 16R8, 16R6, 16R4		MIL R <sub>1</sub> = 390 R <sub>2</sub> = 750	15	30		35	ns
t <sub>H</sub>	Hold Time 16R8, 16R6, 16R4			-10	0		0	ns
t <sub>p</sub>	Clock Period				55		60	ns
t <sub>w</sub>	Clock Width				20		25	ns
f <sub>MAX</sub>	Maximum Frequency				18		16.5	MHz

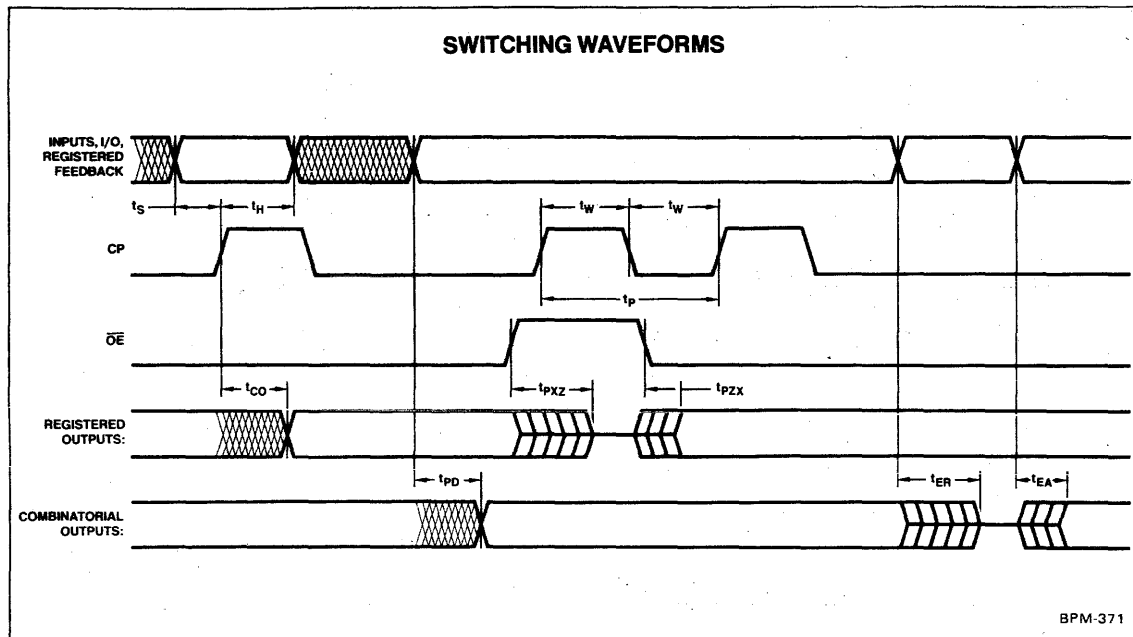
- Notes: 1. Typical limits are at V<sub>CC</sub> = 5.0V and T<sub>A</sub> = 25°C.  
 2. t<sub>PD</sub> is tested with switch S<sub>1</sub> closed and C<sub>L</sub> = 50pF.  
 3. For three-state outputs, output enable times are tested with C<sub>L</sub> = 50pF to the 1.5V level; S<sub>1</sub> is open for high impedance to HIGH tests and closed for high impedance to LOW tests. Output disable times are tested with C<sub>L</sub> = 5pF. HIGH to high impedance tests are made to an output voltage of V<sub>OH</sub> - 0.5V with S<sub>1</sub> open; LOW to high impedance tests are made to the V<sub>OL</sub> + 0.5V level with S<sub>1</sub> closed.



# Advanced Micro Devices

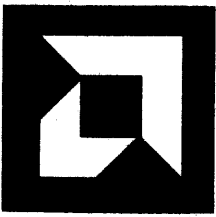


## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements



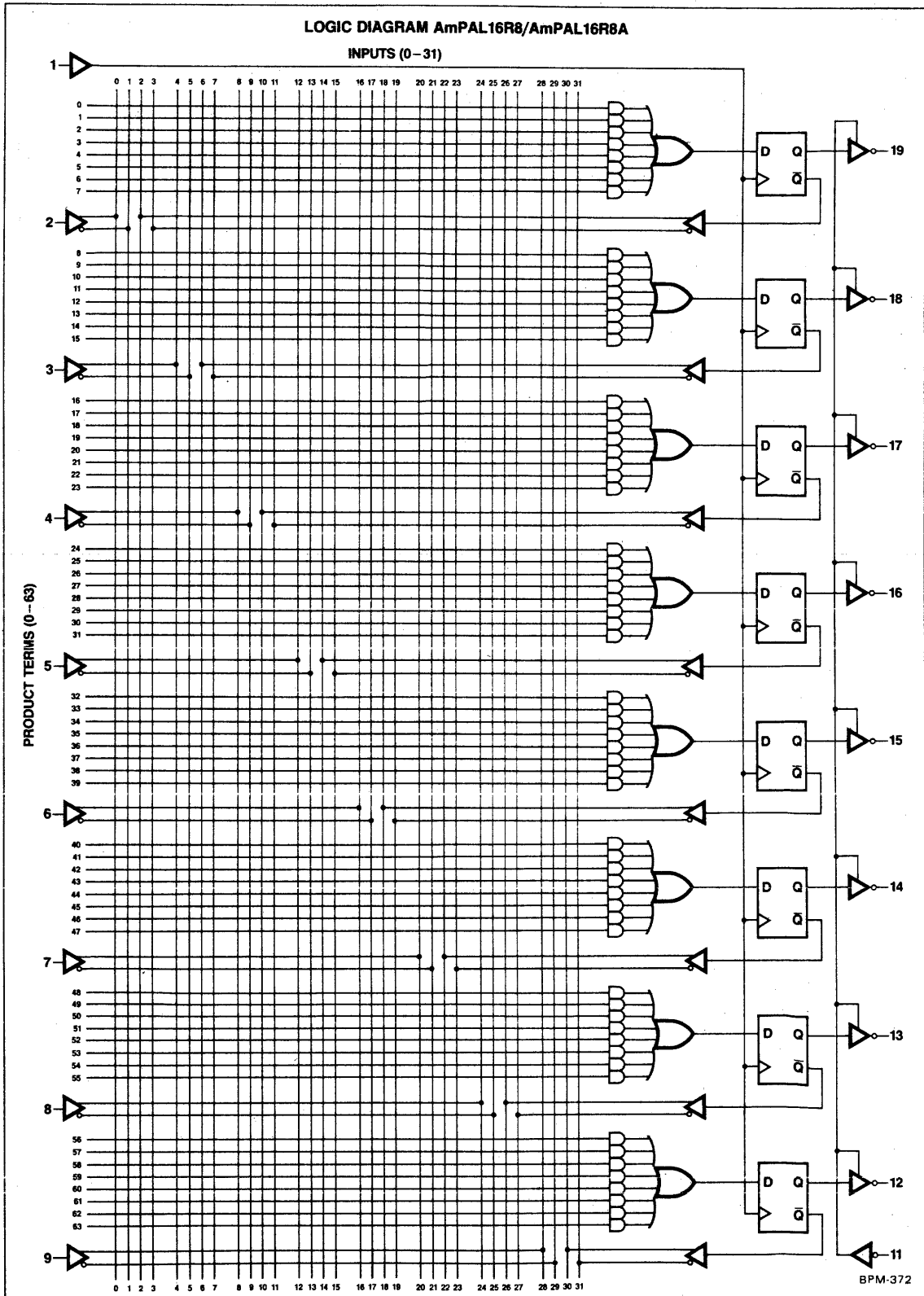
### KEY TO TIMING DIAGRAM

WAVEFORM	INPUTS	OUTPUTS	WAVEFORM	INPUTS	OUTPUTS
	MUST BE STEADY	WILL BE STEADY		DON'T CARE; ANY CHANGE PERMITTED	CHANGING; STATE UNKNOWN
				DOES NOT APPLY	CENTER LINE IS HIGH IMPEDANCE "OFF" STATE



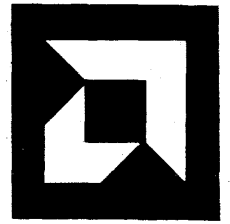
# Advanced Micro Devices

## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOX™ Programmable Array Logic Elements

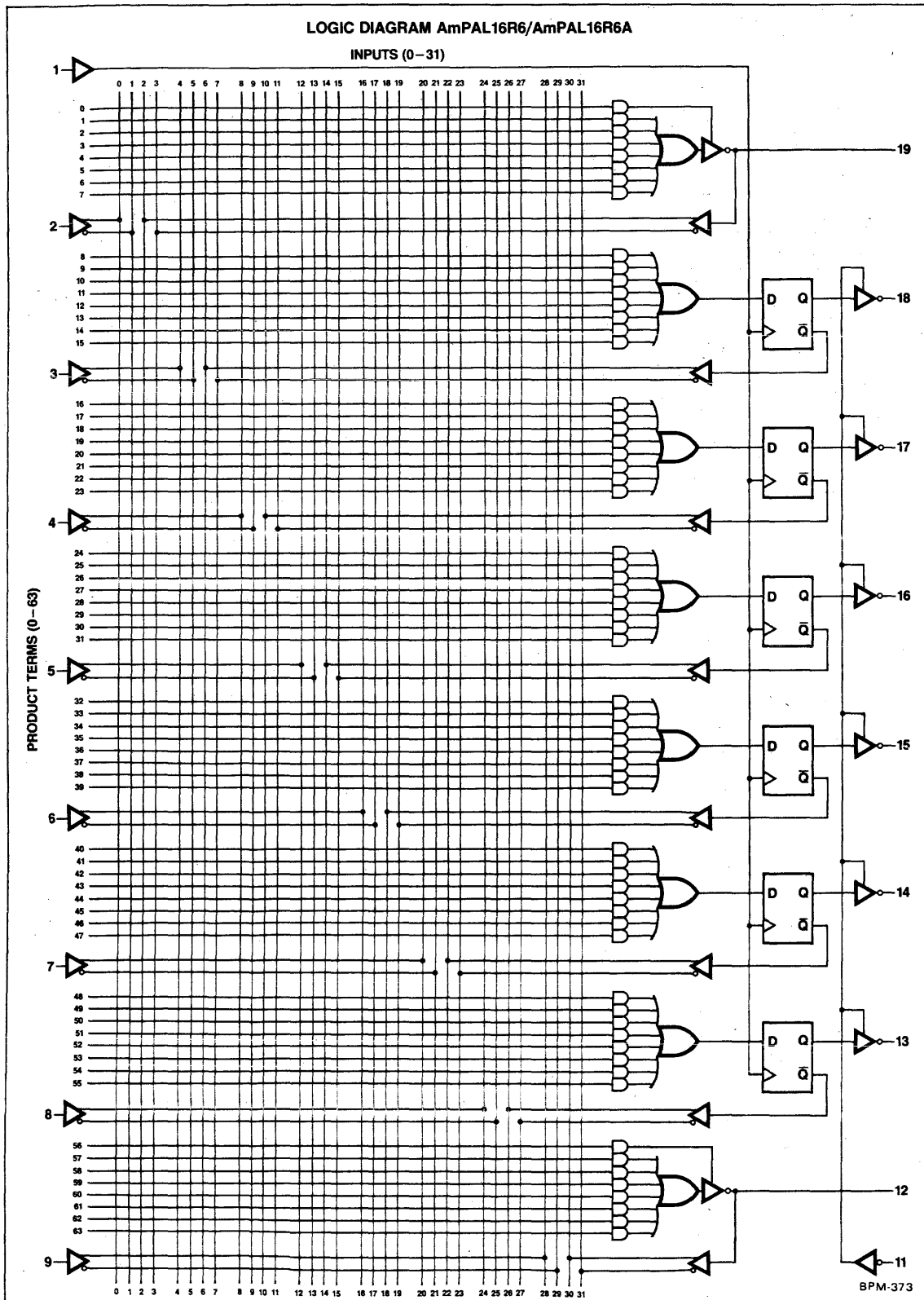


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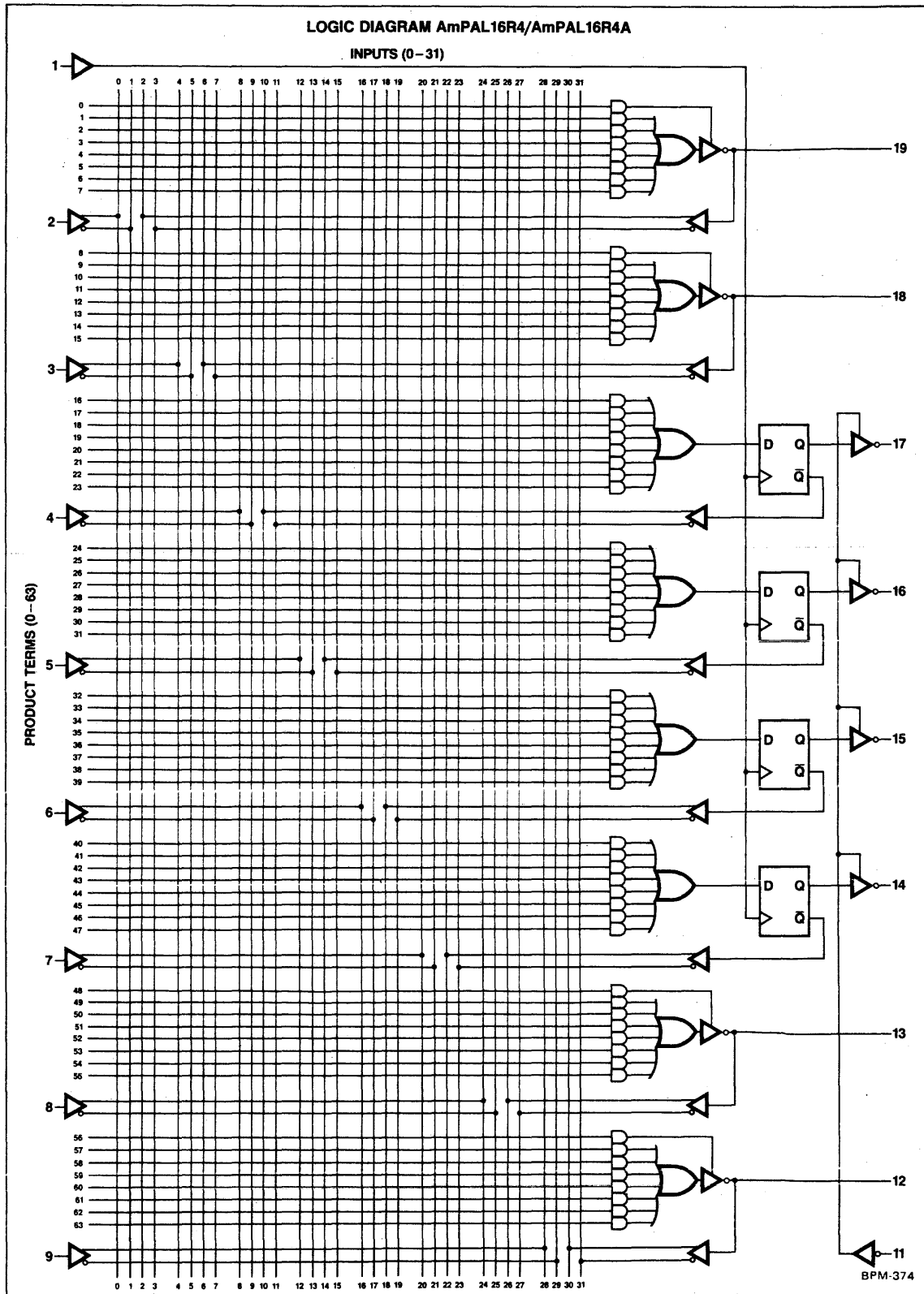
## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements





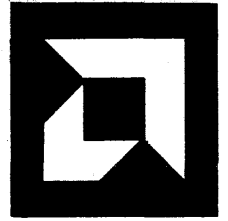
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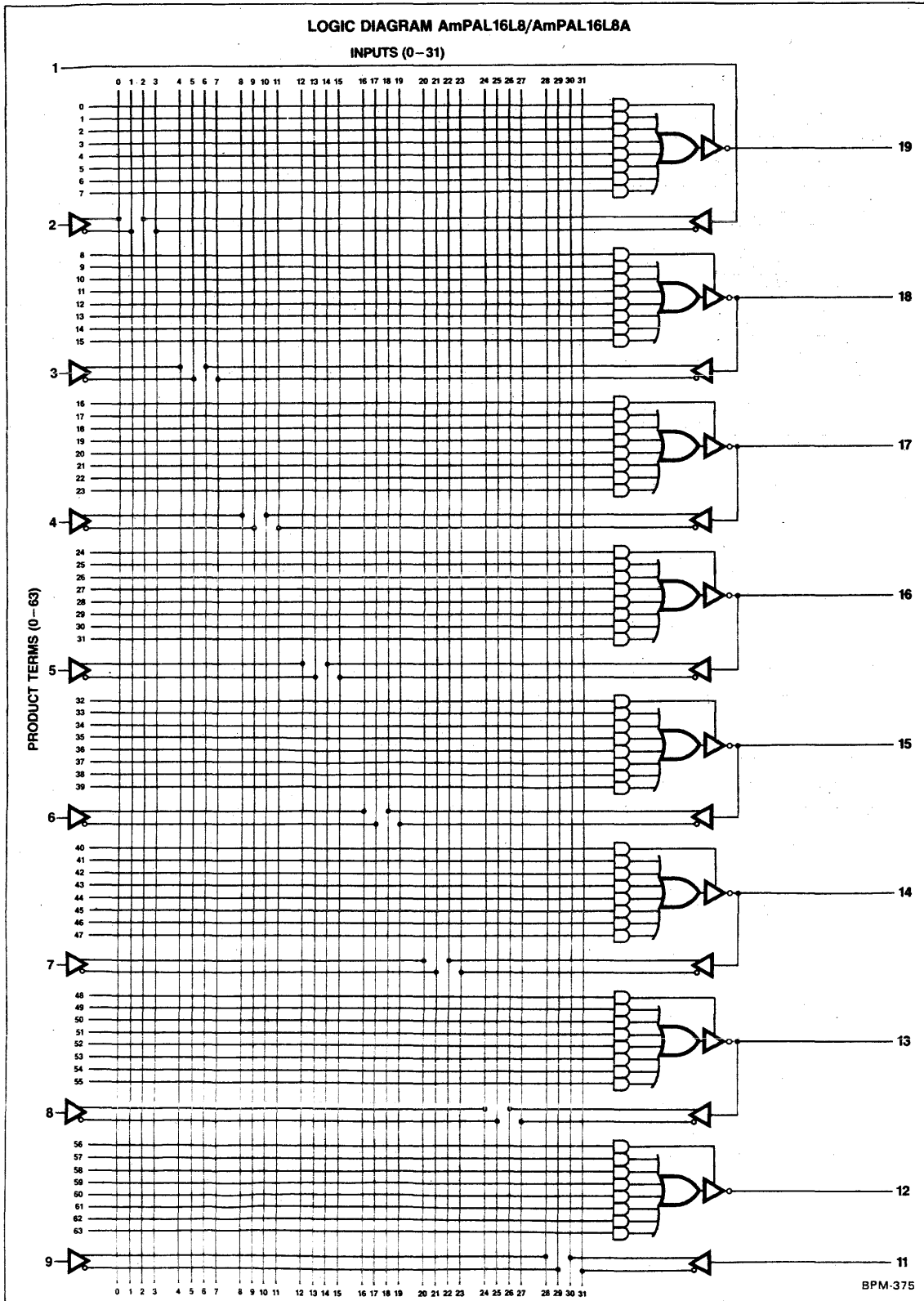


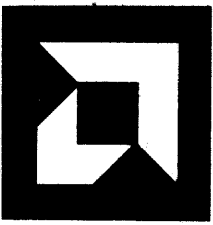
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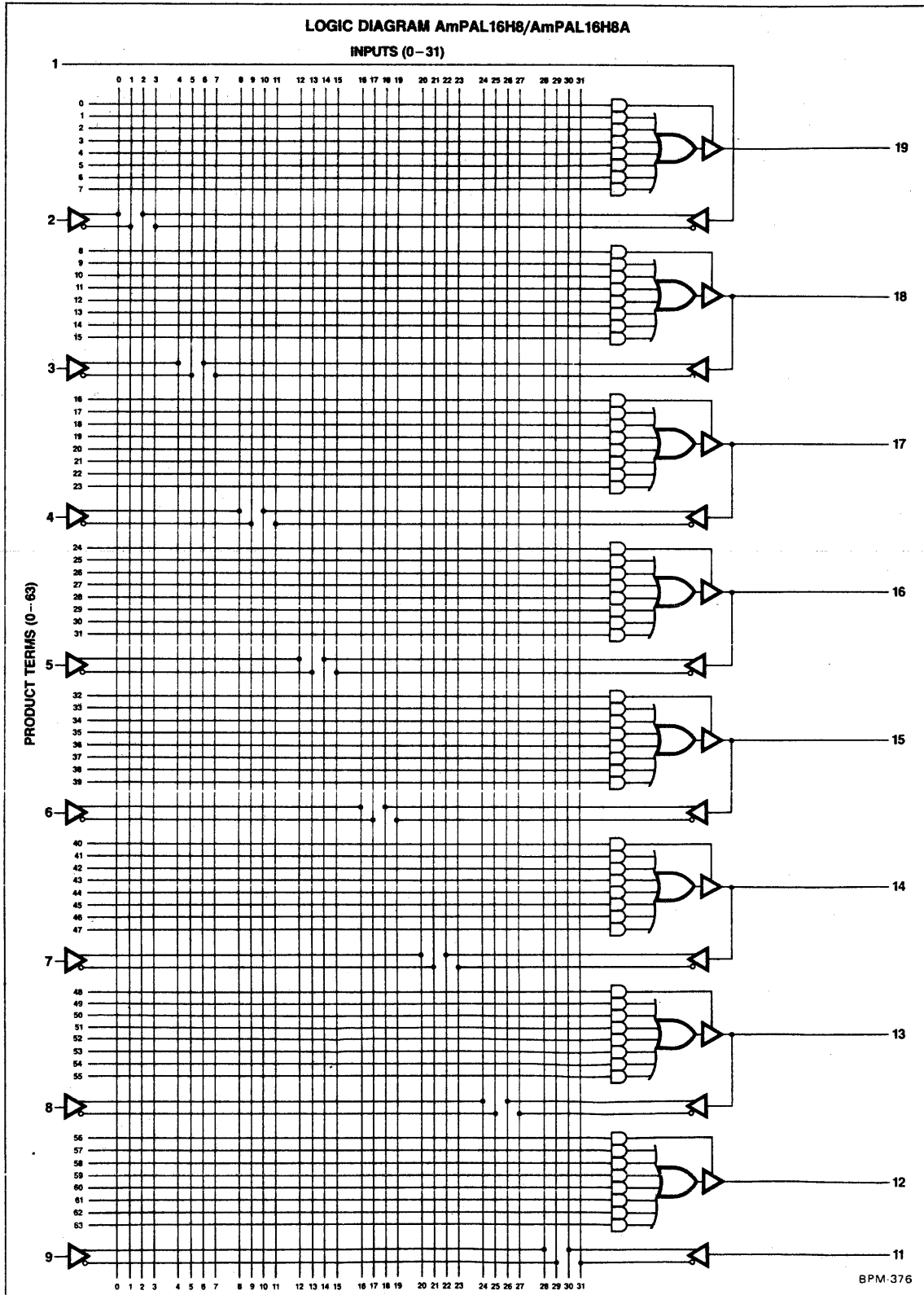
## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOX™ Programmable Array Logic Elements





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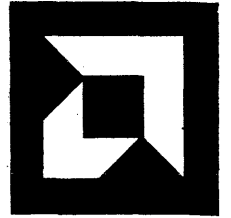
## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements



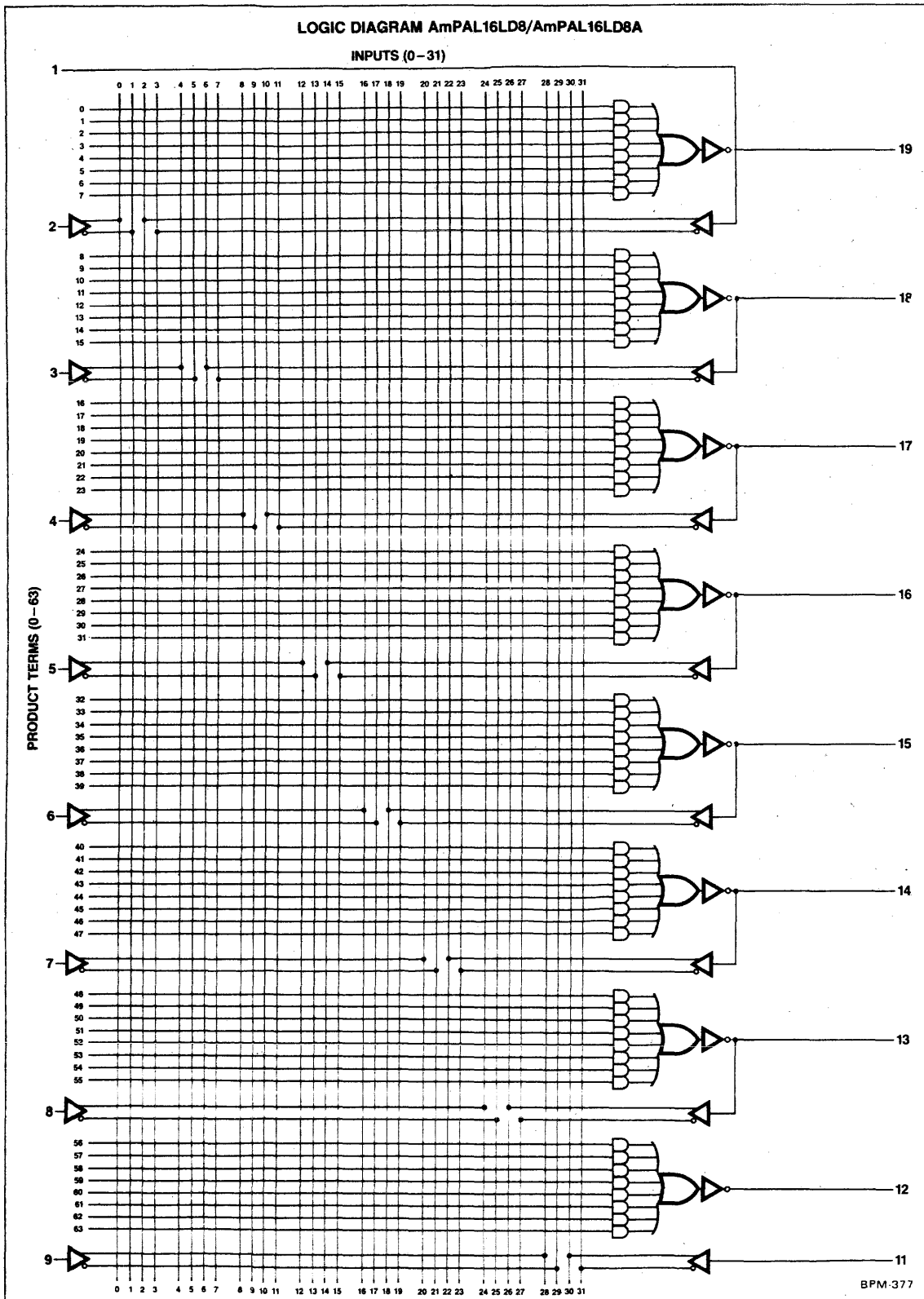
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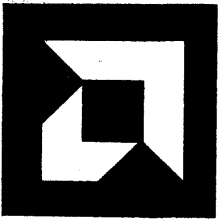
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## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements

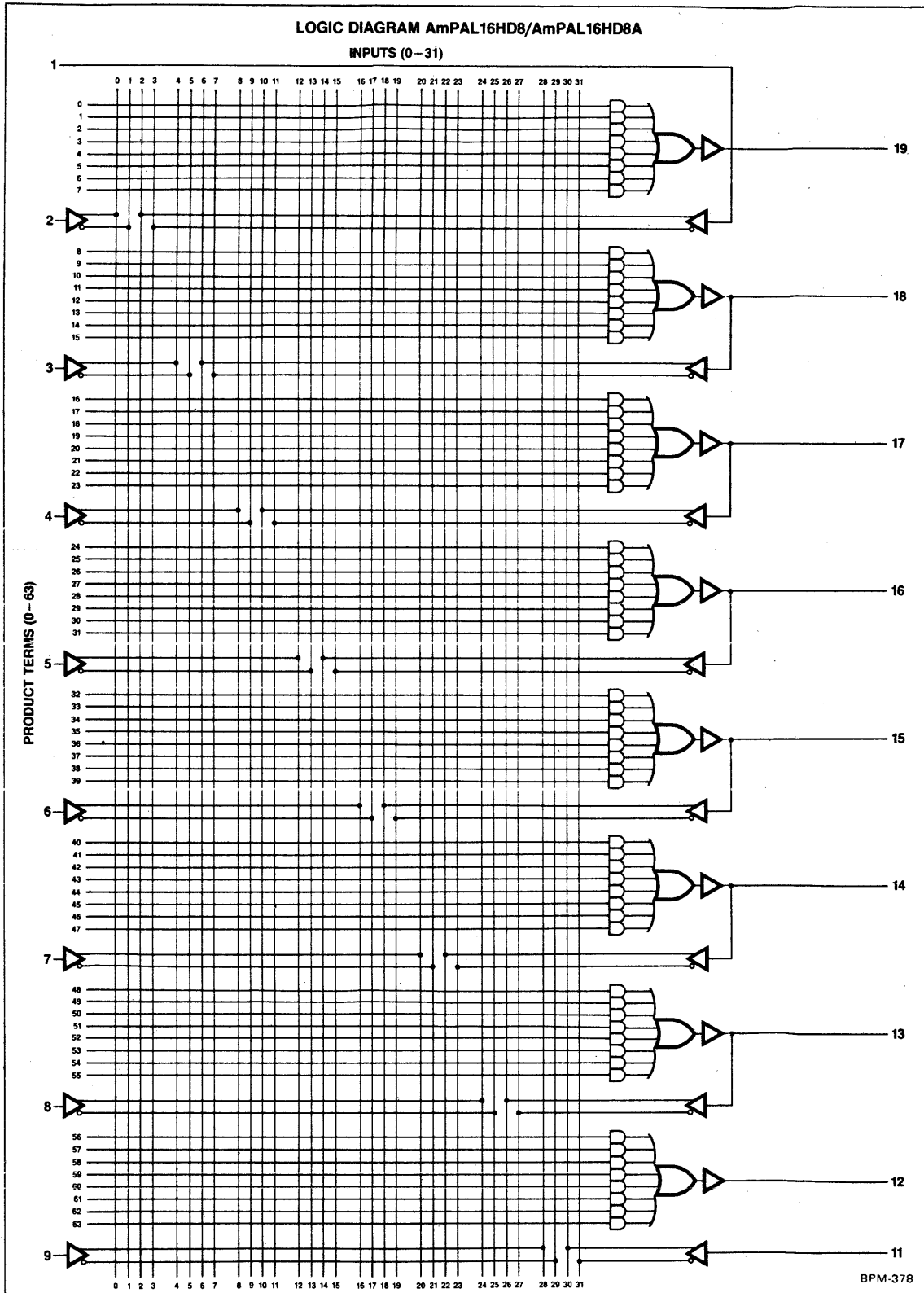






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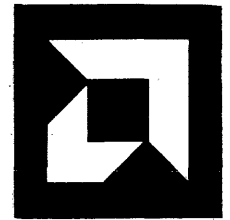
## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements



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## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOX™ Programmable Array Logic Elements

### PROGRAMMING

Each AMD PAL fuse is programmed with a simple sequence of voltages applied to two control pins (1 and 11) and a programming voltage pulse applied to the output under programming. Addressing of the 2048 element fuse array is accomplished with normal TTL levels on eight input pins (five select the input line number and three select the product term number).  $V_{CC}$  is maintained at a normal level throughout the programming and verify cycle – no extra high levels are required.

The necessary sequence of levels for programming any fuse is shown in the Programming Timing Diagram. The address of each fuse in terms of Input Line Number and Product Term Number is defined by the Fuse Address Tables 1 and 2. Current, voltage and timing requirements for each pin are specified in the Programming Parameter Table below.

The 16L8, 16R8, 16R6, 16R4, 16H8, 16LD8 and 16HD8 use identical programming conditions and sequences.

After all programming has been completed, the entire array should be reverified at  $V_{CCL}$  and again at  $V_{CCH}$ . Reverification can be accomplished by reading all eight outputs in parallel rather than one at a time. The array fuse verification cycle checks that

the correct array fuses have been blown and can be sensed by the outputs.

AMD PALs have been designed with many internal test features that are used to assure high programming yield and correct logical operation for a correctly programmed part.

An additional fuse is provided on each AMD PAL circuit to prevent unauthorized copying of AMD PAL fuse patterns when design security is desired. Blowing the security fuse blocks entry to the fuse pattern verify mode.

To blow the security fuse:

1. Power up part to  $V_{CCP}$
2. Raise Pin 5 to  $V_{HH}$ .
3. Pulse Pin 11 from ground to  $V_{OP}$  for a 50 $\mu$ sec duration.
4. Perform a normal end-of-programming verify cycle at  $V_{CCL}$  and  $V_{CCH}$ . All fuse locations should be sensed as blown if the security fuse has been successfully blown.

Note that parts with the security fuse blown may not be returned as programming rejects.

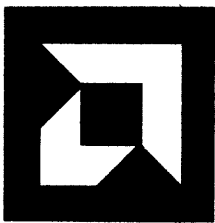
AMD PALs normally have high programming yields (>98%). Programming yield losses are frequently due to poor socket contact, equipment out of calibration or improperly used.

### PROGRAMMING PARAMETERS $T_A = 25^\circ\text{C}$

Parameters	Description	Min	Typ	Max	Units	
$V_{HH}$	Control Pin Extra High Level	Pin 1 @ 10–40mA	10	11	12	Volts
		Pin 11 @ 10–40mA	10	11	12	
$V_{OP}$	Program Voltage Pins 12–19 @ 15–200mA	18	20	22	Volts	
$V_{IHP}$	Input High Level During Programming and Verify	2.4	5	5.5	Volts	
$V_{ILP}$	Input Low Level During Programming and Verify	0.0	0.3	0.5	Volts	
$V_{CCP}$	$V_{CC}$ During Programming @ $I_{CC} = 50\text{--}200\text{mA}$	5	5.2	5.5	Volts	
$V_{CCL}$	$V_{CC}$ During First Pass Verification @ $I_{CC} = 50\text{--}200\text{mA}$	4.1	4.3	4.5	Volts	
$V_{CCH}$	$V_{CC}$ During Second Pass Verification @ $I_{CC} = 50\text{--}200\text{mA}$	5.4	5.7	6.0	Volts	
$V_{Blown}$	Successful Blown Fuse Sense Level @ Output	16L8, 16R8, 16R6, 16R4, 16LD8 16L8A, 16R8A, 16R6A, 16R4A, 16LD8A		0.3	0.5	Volts
		16H8, 16HD8, 16H8A, 16HD8A	2.4	3		
$dV_{OP}/dt$	Rate of Output Voltage Change	20		250	$V/\mu\text{sec}$	
$dV_{11}/dt$	Rate of Fusing Enable Voltage Change (Pin 11 Rising Edge)	100		1000	$V/\mu\text{sec}$	
$t_p$	Fusing Time First Attempt	40	50	100	$\mu\text{sec}$	
	Subsequent Attempts	4	5	10	msec	
$t_D$	Delays Between Various Level Changes	100	200	1000	ns	
$t_V$	Period During which Output is Sensed for $V_{Blown}$ Level			500	ns	
$V_{ONP}$	Pull-Up Voltage On Outputs Not Being Programmed	$V_{CCP} - 0.3$	$V_{CCP}$	$V_{CCP} + 0.3$	Volts	
R	Pull-Up Resistor On Outputs Not Being Programmed	1.9	2	2.1	K $\Omega$	

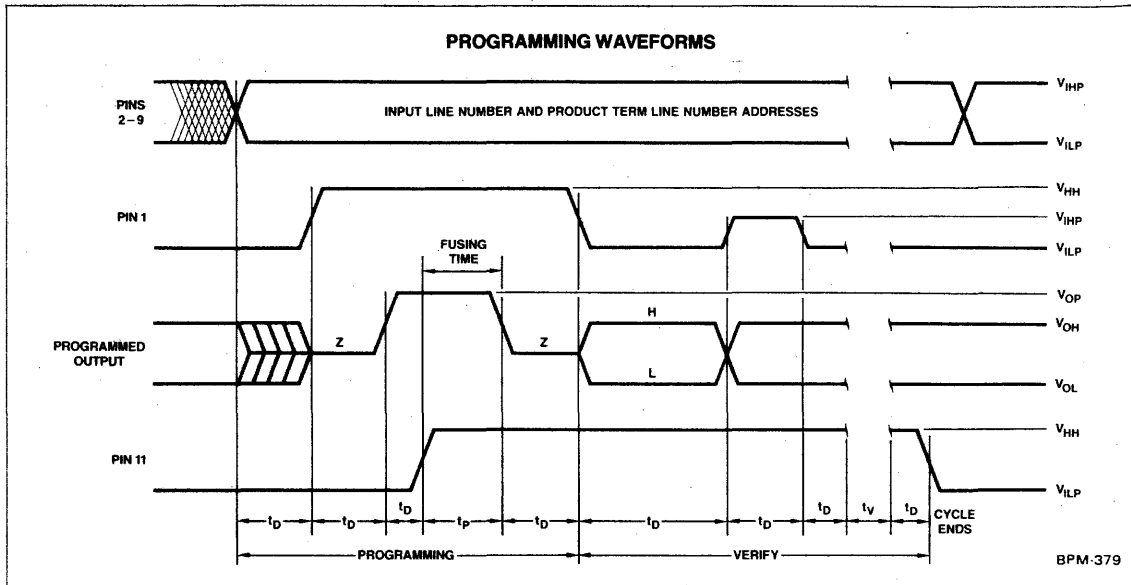
### AMD PAL PROGRAMMING EQUIPMENT INFORMATION

Source and Location	Data I/O 10525 Willows Rd. N.E. Redmond, WA 98052	Kontron Electronics, Inc. 630 Price Avenue Redwood City, CA 94063	Stag Microsystems 528-5 Weddel Drive Sunnyvale, CA 94086
Programmer Model(s)	Model-100, 29, 19 or 17	Model-MPP-80S or EPP80	Model-PPX
AMD PAL Personality Module	Logicpak 950-1942-001	MOD-33	PPM2200
Socket Adapter	715-1947-003	SA37	Am202S



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## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements

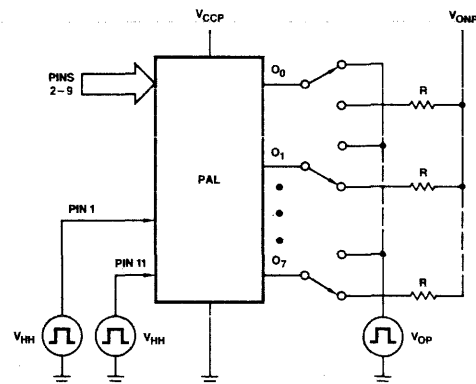


**TABLE 1. INPUT ADDRESSING**

Input Line Number	Input Line Number Address Pin States				
	9	8	7	6	5
0	L	L	L	L	L
1	L	L	L	L	H
2	L	L	L	L	L
3	L	L	L	L	H
4	L	L	L	H	L
5	L	L	L	H	H
6	L	L	L	H	H
7	L	L	L	H	H
8	L	L	H	L	L
9	L	L	H	L	L
10	L	L	H	L	H
11	L	L	H	L	H
12	L	L	H	L	L
13	L	L	H	L	H
14	L	L	H	L	L
15	L	L	H	L	H
16	H	L	L	L	L
17	H	L	L	L	L
18	H	L	L	L	H
19	H	L	L	L	H
20	H	L	L	L	L
21	H	L	L	L	H
22	H	L	L	L	L
23	H	L	L	L	H
24	H	L	L	L	L
25	H	L	L	L	H
26	H	L	L	L	L
27	H	L	L	L	H
28	H	L	L	L	L
29	H	L	L	L	H
30	H	L	L	L	L
31	H	L	L	L	H

L =  $V_{ILP}$   
H =  $V_{IHP}$

**SIMPLIFIED PROGRAMMING DIAGRAM**



**TABLE 2. PRODUCT TERM ADDRESSING**

Product Term Line Number								Product Term Select Address Pin		
0	8	16	24	32	40	48	56	4	3	2
1	9	17	25	33	41	49	57	L	L	L
2	10	18	26	34	42	50	58	L	L	H
3	11	19	27	35	43	51	59	L	H	L
4	12	20	28	36	44	52	60	H	L	L
5	13	21	29	37	45	53	61	H	L	H
6	14	22	30	38	46	54	62	H	H	L
7	15	23	31	39	47	55	63	H	H	H

Pin 19   Pin 18   Pin 17   Pin 16   Pin 15   Pin 14   Pin 13   Pin 12

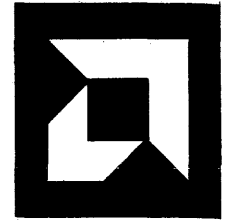
Programming Access and Verify Pin

L =  $V_{ILP}$   
H =  $V_{IHP}$

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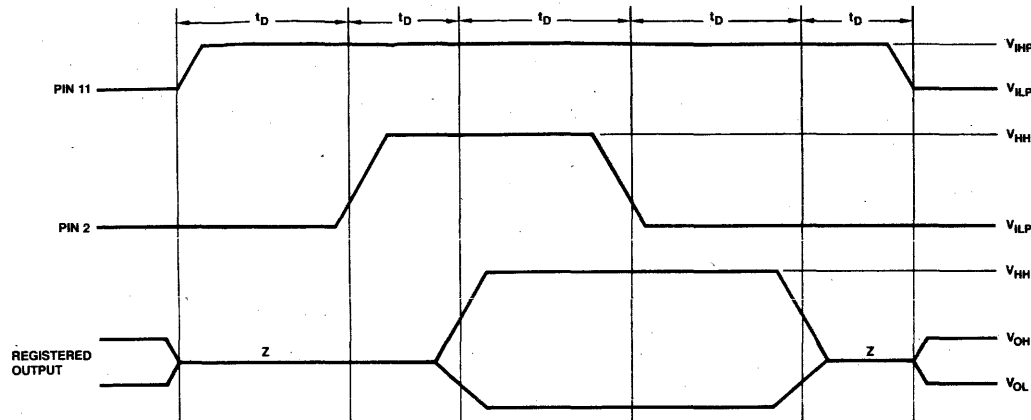
## BIPOLAR LSI AND SUPPORT PRODUCTS 20-Pin IMOXTM Programmable Array Logic Elements

### PRELOAD OF REGISTERED OUTPUTS

AMD PAL registered outputs are designed with extra circuitry to allow loading each register asynchronously to either a HIGH or

LOW state. This feature simplifies testing since any initial state for the registers can be set to optimize test sequencing.

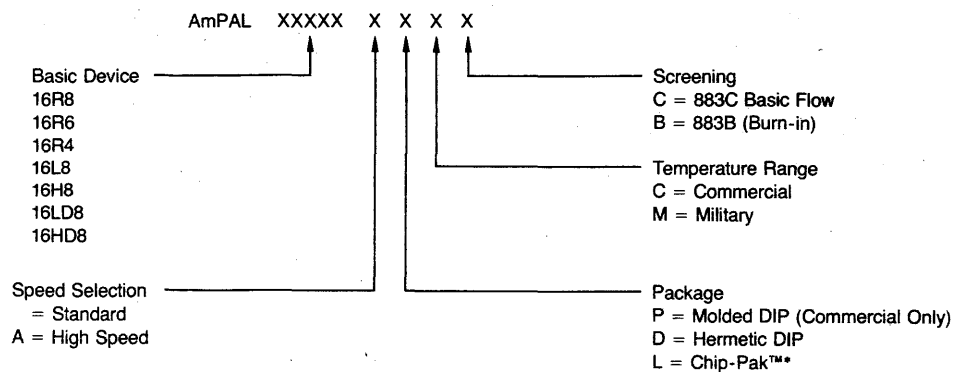
The pin levels and timing necessary to perform the PRELOAD function are detailed below:



Level forced on registered output pin during preload cycle	Output state at the output pin after cycle
V <sub>HH</sub>	HIGH
0V to V <sub>CCH</sub> or OPEN	LOW

BPM-381

### ORDERING INFORMATION

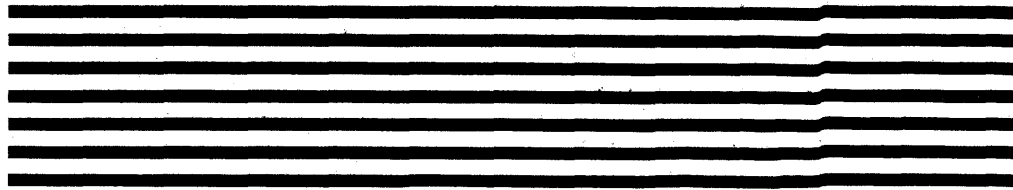


\*Chip-Paks are rated at maximum case temperature only.

Chip-Pak is a trademark of Advanced Micro Devices, Inc.

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## Spectrum of Custom Solutions

No other company can match AMI's track record in developing state-of-the-art custom VLSI products. With more than 2000 custom devices designed and manufactured since 1966, AMI has more experience than any other integrated circuit company in building a wide variety of custom microcircuits.

AMI not only has the experience, but the design engineering organization and the advanced production and testing facilities to produce the highest quality MOS/VLSI circuits. And because AMI also offers standard memory, microprocessor, telecommunication and consumer products and the widest variety of custom VLSI processes in the industry, we're able to be objective in helping customers determine their most cost effective approach.

### The Advantages of Custom Circuits

Since a single custom MOS/VLSI chip can replace expensive electromechanical devices, discrete logic components, or less efficient general purpose LSI circuits, it offers a number of benefits not available with standard logic.

**Custom circuits save money.** Grouping functions onto a single chip lowers production and inventory costs dramatically. That reduces your product manufacturing costs as well.

**Custom circuits are more reliable.** Putting a complete system on a chip trims component count, improving both product reliability and production yields. Rework, repair and replacements are minimized.

**Custom circuits reduce space and power requirements.** Fewer components means both space and power requirements are reduced.

**Custom circuits offer superior performance.** Since the circuit is designed to your requirements, features and functions can be incorporated which are not available in general purpose chips. Special tailoring reduces test requirements as well.

**Custom circuits offer proprietary protection.** Being tailored exactly to your requirements, a custom circuit cannot be easily duplicated. This can help put you ahead — and keep you ahead — of your competition.

### The Spectrum of Solutions

The decision to use a custom circuit depends on your system design requirements — such things as complexity, features, size and power limitations. But no longer is your custom decision limited by low volume or short development time — not when you come to AMI.

AMI has a full spectrum of custom solutions to assure you get that solution which meets your system performance and time-to-market requirements at the lowest possible cost.

AMI's spectrum of solutions bridges the total span of volume, timing and interface needs of our customers. From semi-custom designs, to full custom design — somewhere

on the spectrum, your development time and volume requirements can be met. For customers who already have their designs, AMI can provide custom fabrication for the customer's tooling. We will even teach custom design if that's what our customers need. And we can even go a step further and license the technology for a customer to set up his own fabrication capability. No other company offers such a spectrum of solutions. And no other company has more experience at helping you pick the best solution for your needs.

### Semi-Custom Gate Arrays

AMI offers both gate arrays and standard cell design methods for semi-custom circuit development.

Gate arrays are the best solution for circuits of moderate complexity in low-to-medium volume applications or where the shortest possible development time is required. AMI offers both gate arrays and standard cell design methods for semi-custom circuit development.

AMI CMOS semi-custom gate arrays are standard logic layouts of everything except the final metal interconnect pattern. Since only the final pattern needs to be developed to customize your circuits, both development time spans and development costs are dramatically reduced. Because wafers containing arrays are preprocessed and inventoried, production lead times are short. Gate arrays are especially attractive for applications requiring circuit volumes from 1,000 to 50,000 units per year.

For more details on AMI's gate arrays, refer to the "Semi-Custom" section.

### Standard Cell Custom

Standard cells are custom circuits which are designed from computer stored modular cells. The computer assembles the cells into a collection of functional blocks to form a custom circuit. Since standard cells utilize predesigned cells, development time is reduced dramatically and development costs are cut 30 to 50 percent over conventional custom design. Circuit size is likely to be slightly larger than a conventional custom circuit, so they are most appropriate where rapid development is more important than minimal size. Standard cells are cost effective in volume levels beginning around 10,000 circuits.

For more details on AMI standard cells refer to the "Semi-Custom" section.

### Optimized Custom Design

Where end product volume is high — beyond 50,000 units per year — or where special requirements for lowest power, minimal space or highest performance exist, the solution is likely to be conventional custom design. By optimizing circuit elements and layout for a specific part, die size is substantially smaller than using semi-custom design methods. In high volume applications, a smaller die size results in lower unit cost to the customer.



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## Spectrum of Custom Solutions

In addition, custom designs allow you to combine logic elements, memory, and analog circuits in a single device. This design flexibility is not available in gate arrays and only available to a limited extent in standard cells.

Instrumental in the design of custom circuits is our Symbolic Interactive Design System (SIDS). The design is done primarily with SIDS where a layout designer works with symbols directly at a large screen alphanumeric color CRT. After the SIDS circuit design has been completed and verified, the symbols are converted to polygons and a 10X reticle tape is prepared.

SIDS uses on-line, real-time design rule checking capability to isolate design rule errors in the layout. This allows immediate correction which greatly reduces the development span time.

Also a nodal trace function permits a designer to trace and highlight a given electrical node. In this way, the designer can manually insure that the node is connected as specified in the master logic description.

Full background real-time design rule checking on windows, cells, and chips is supported, as is full background continuity checking against the master logic description. This eliminates the delay from digitizing and batch processed computer checking of circuits for accuracy.

With SIDS, error correction, circuit modification and area relocations take only minutes. That significantly reduces design cycle time and development costs.

Computer-aided hand-drawn layouts are used to reduce extremely complex circuits to the absolute smallest size. Development time and costs are higher, but in certain cases, size or complexity requirements may require the hand-drawn approach.

### Customer Owned Tooling

In many cases AMI customers design their own circuits, either through an internal design center, a design house, or another vendor. In these cases, AMI can provide custom circuit fabrication from customer owned tooling. In conjunction with a silicon foundry service, AMI's Customer Owned Tooling (COT) group can offer the customer design support in the form of design rules, standard cells, a three phase development program, test program generation and general technical support. In the event of design problems, AMI's COT group utilizes the design expertise of the custom or semi-custom group for technical assistance.

Once the circuit is designed, the customer has the option of transferring the circuit tooling to AMI in PG tape, data base tape or working plate form. Close cooperation between the customer and AMI's COT Group during an established three phase prototype development program simplifies circuit debug and reduces risk of design error.

Drawing on AMI's extensive manufacturing ability, COT offers a wide range of process variations, package types, and test capabilities. Unlike most of AMI's competition in

the silicon foundry market, AMI has internal photolithography (mask making) and assembly capability. This internal capability avoids the high costs and quality risks of subcontracting two very important steps in custom IC manufacturing.

Once the prototype development program is completed, AMI can ship as few as five to as many as thousands of wafers per week to the customer. AMI can ship tested die, fully tested packaged devices and devices meeting military 883-B standards.

AMI's Customer Owned Tooling group is the only vendor in the silicon foundry market that offers design support, full service manufacturing capability, flexibility and experience.

### Joint Development Ventures

Through a Joint Development Team (JDV) we can teach a customer to design his own MOS/VLSI circuits. The JDV is a combination of technically skilled people from the partner company and AMI who function as a design group concentrating on the customer's products alone. The JDV partner brings his system design staff and AMI brings the MOS/VLSI staff and its design technology. The partner becomes part of an in-house AMI design group. The end result is a design capability for the partner company for circuits that AMI will fabricate.

If the customer wants to go beyond designing his own circuits to operating his own manufacturing/pilot line, AMI will license the necessary technology in those situations where a long-term business relationship can be established between the partner company and AMI.

### AMI Provides Leading CAD Technology

At almost all levels of the spectrum, computer-aided design (CAD) software and hardware aids are employed to assure correctness of design each step of the way and to shorten design spans reducing customer risk and lowering design cost. Highly efficient programs have been implemented to assist in logic design and simulation, layout planning, switched capacitor analysis routines and symbolic interactive design layout, to name just a few.

#### Hardware design aids include:

- On-site Burroughs 7760 computer with multiprocessing capability.
- Computer terminals built around a Prime computer and engineering design facilities which tie into the on-site 7760 and time-sharing services.
- Computervision interactive graphics system which provide on-line generation and editing of composite drawings; includes drafting surfaces and CRT displays.
- Calma graphics system for both production digitizing and on-line changes.
- Calma GDS-11 high speed electrostatic plotter.



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## Spectrum of Custom Solutions

- High speed, high resolution Electromask 9-track pattern generator.

### Software design aids include:

#### Logic Design

- **Register Transfer Language (RTL) Simulation** — Provides a system behavior description to define instruction sets, optimize data paths, control hardware algorithm design and establish register designs.
- **Glide** — Permits user to design layout, simulate, generate patterns and develop test programs for logic arrays.
- **Path Analysis Program (PATH)** — Permits gross logic checks to be made before design, and final logic checks from the ultimate design.
- **Logic Simulator (SIMAD)** — (SIMulator with Assignable Delays) simulates logic network behavior for design verification and propagation delays.
- **Programmable Logic Array Design Aids (PLAID)** — Uses state tables and Boolean equations to generate the optimum physical structure for random logic designs.
- **Block Oriented Logic Translator (BOLT)** — A logic description compiler that generates a common data base used by SIDS, SIMAD, LPA, continuity check, PATH and CIPAR.
- **Design Rule Checking (DRC)**
- **Trace and Continuity Checking**

#### Circuit Design

- **Circuit Simulator (ASPEC)** — Analyzes DC operation, DC transfer functions, time domain or transients and frequency domain or small signal AC characteristics.
- **Pole Zero Analysis (PZSLIC)** — Program analyzes the frequency domain of linear integrated circuits.
- **Switched Capacitor Analysis Routine (SCAR)** — Analyzes switched capacitor filter designs for telecommunications and other analog circuits.
- **Data Analysis Program (DAP)** — Analyzes data from circuit fabrication to maintain the parameters of circuit designs.

#### Mask Design

- **Layout Planning Aid (LPA)** — Lays out the chip plan and interconnection between functional blocks of an integrated circuit.
- **Symbolic Interactive Design System (SIDS)** — Permits a layout designer to work directly with a computer to lay out and check a circuit on a CRT screen, dramatically shortening layout time requirements.
- **Circuit Interactive Place and Route (CIPAR)** — Automatically creates error-free mask designs in extremely short time spans.

#### Test Generation

AMI utilizes numerous software programs to generate test programs for integrated circuits. All serve to reduce the

time needed to develop test programs to meet customer specifications.

### Digital and Analog Combinations

AMI is a leading innovator in combining digital and analog functions on a single chip. We can combine any of the functions below into an optimum circuit configuration to meet your needs. Unique combinations of these functions are already used in many applications in the communications, consumer, and industrial marketplace.

DIGITAL	ANALOG
PLA	OP AMP
ALU	Oscillator
Inverter	Comparator
RAM and ROM	Voltage Reference
Shift Register	A/D and D/A Converters
Interface Driver	Switched Capacitor Filters
Automatic Power Down	Programmable Power Down
	Phase Locked Loops

### State of the Art Packaging

AMI's packaging capability spans a broad spectrum, beginning with plastic, ceramic and CERDIP and going on to chip carriers, die bonding to PC boards and, most recently, mini-flat packs. As well as being a leader in plastic packaging for the high volume, low cost consumer industry, AMI's high reliability plastic packages and chip carriers are accepted under the stringent requirements in the Telecom and Automotive industries. As many industry segments move toward space-saving packages, AMI remains in the forefront in packaging using chip carriers. AMI now is developing a family of mini-flat packs which are a plastic alternative to a chip carrier.

### AMI Delivers Quality

AMI quality controls for in-process wafer inspection and final assembly and test are the best in the industry. Our care in fabrication, assembly and test mean that you get products that meet your specifications for reliability. Because over 70 percent of our total production is custom, we perform many checks routinely that would only be done on special orders and at additional cost by other manufacturers. In fact, our own in-house standards are tougher than most of our customers require. Most importantly, AMI is committed to making sure that everything we do is done right, every time we do it.

### The Industry's Highest Standard

AMI has consistently pursued product excellence and has reached for higher quality levels in finished products shipped. Circuits are inspected to 0.04% AQL or your specifications, whichever is more stringent.

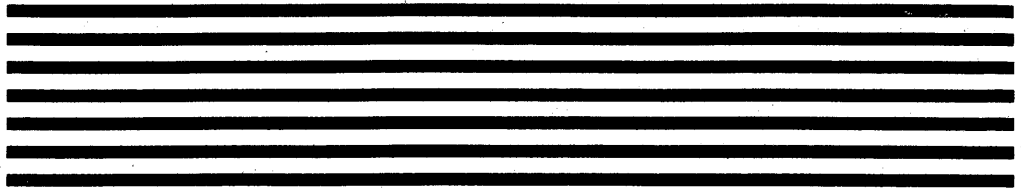
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## Spectrum of Custom Solutions

This 0.04% AQL can put you in a superior competitive position. Your incoming test and assembly costs come down since there is less reworking on the line. And your customers receive a more reliable product.

### Quality Checks

Among the routine quality controls exercised over every product at AMI are:

- Full logic design checks against system specifications
- Circuit simulation to verify performance against objectives
- Working plates check on automatic checkers
- Automated mask fabrication checks

- In-process wafer fabrication checks
- Wafer sort tests
- 100% optical inspection at dicing
- 100% die attach checking
- 100% lead bonding inspection prior to package sealing
- Seal checks, fine and gross leak tests
- Final digital and analog tests
- Customer specified environmental tests

Meticulous in-process checks are performed on design and workmanship at every step, to ensure a full manufacturable device. In manufacture, lot process and yield data are captured and examined as a matter of routine.

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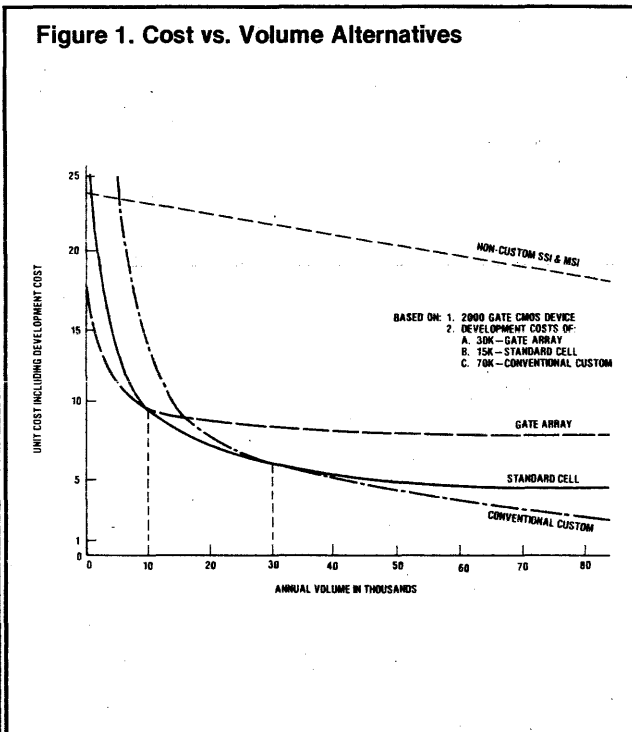




## Semi-Custom Capabilities

### Overview

As the semiconductor industry has marched into the new era of VLSI, a new market has appeared—fast turn custom or, as it is now called, semi-custom. AMI, a leader in custom MOS since 1966, is also a leader in this new semi-custom market. AMI has introduced CAD software and hardware tools to allow customers to design, simulate, and layout circuits using AMI gate array and standard cell families. Figure 1 shows the economic tradeoffs between gate array, standard cell, and full custom, all of which are offered by AMI.



### Gate Arrays

The simplest semi-custom ICs are gate arrays. As the name implies, a gate array consists of uncommitted component matrices of transistors (usually P- and N-type for CMOS) that allow user-defined interconnections through a single or double layer of metal. Since arrays employ fixed component locations and geometries, AMI can process the wafers up to the metallization stage and inventory the wafers for future customization. Thus gate arrays look like late mask programmable ROMs and benefit from this large-volume production because they appear to be a standard product. Thus AMI can offer them at an economical price and with fast prototyping and production turn on spans.

### Standard Cells

The second semi-custom product group is standard cells.

Standard cells employ fully customized process/mask sets and must pass through all process steps before a user-specified circuit is completed. To design such chips, AMI customers use precharacterized functional cells from our cell libraries. Placing and routing the cells is done on AMI computers using specially developed software. Standard cell designs usually result in smaller chips since only the component structures required for the user-specified circuit are included, thus chips designed with cells are less expensive than gate array designed chips.

The key to success in this new market is flexibility. Flexibility to the user entails: low risk circuit implementation, short development span, lower development cost, lower piece part cost (over TTL implementations), easy to change or modify, enhanced product features, etc. For the manufacturer, flexibility means: ease of manufacture, economies of scale, and easy interface with customers. One last point: AMI offers the user the opportunity to migrate at a low cost, from a gate array to a standard cell (or possibly full custom) to further enhance his/her product. By using analog cells, significant advances in chip function integration are at the user's disposal.

In addition, AMI offers a wide selection of packages to meet specific user needs. AMI offers the CAD tools needed to work in the new market. AMI also offers the training required to move customers quickly and easily into this new technology.

### 2-Micron Products

AMI is developing 2-micron CMOS technology to support the next generation of semi-custom products, in both gate arrays and standard cells. These products will offer size and performance improvements of up to 50% from their 3-micron counterparts.

Introduction of the first 2-micron gate array family is planned for second quarter 1984 and is expected to offer capabilities of greater than 8000 gates.

For detailed information on Gate Arrays and Standard Cells consult the following seven pages.



## Gate Arrays

### Gate Arrays

#### Features

- Arrays of Uncommitted CMOS Transistors Programmed by Metal Layer Interconnect to Implement Arbitrary Digital Logic Functions
- Multiple Developmental Interfaces: AMI or Customer Designed
- Three Array Families—5-Micron Single Metal CMOS, 3-Micron Single Metal CMOS, and 3-Micron Double Metal Versions
- Multiple Array Configurations—From 300 to 1260 Gates for 5-Micron Devices, and 500 to 5000 Gates for 3-Micron Devices
- Quick Turn Prototypes and Short Production Turn-On Time
- Economical Semi-Custom Approach for Low-to-Medium Production Volume Requirements
- Advanced Oxide-Isolated Silicon Gate CMOS Technology
- High Performance—2 to 3ns Typical Gate Delay for 3-Micron Devices
- Broad Power Supply Range—3V to 12V ( $\pm 10\%$ )
- TTL or CMOS Compatible I/O
- Up to 134 I/O Connections
- Numerous Package Options
- Full Military Temperature Range ( $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ ) and MIL-STD-883 Class B Screening Available

### General Description

AMI's gate array products consist of arrays of CMOS devices whose interconnections are initially unspecified. By "programming" interconnect at the metal layer mask level, virtually arbitrary configurations of digital logic can be realized in an LSI implementation.

AMI gate array designs are based on topological cells—i.e., groups of uncommitted silicon-gate N-Channel and P-Channel transistors—that are placed at regular intervals along the X and Y axes of the chip with intervening polysilicon underpasses. Pads, input protection circuitry, and uncommitted output drivers are placed around the periphery.

Compared to SSI/MSI logic implementations, AMI's gate array approach offers lower system cost and, in addition, all the benefits of CMOS LSI. The lower system cost is due to significant reductions in component count, board area and power consumption. Product reliability, a strong function of component count, is thereby greatly enhanced. And compared to fully custom LSI circuits, the gate array offers several advantages: low development cost; shorter development time; shorter production turn-on time; and low unit costs for small to moderate production volumes.

AMI's CMOS gate arrays are offered in three families: the 5-micron UA series, the 3-micron single metal GA series, and the 3-micron double metal GA-D series. The 5-micron UA series has been in production since 1980 and well over one hundred circuits have been produced in that technology. The 3-micron GA and GA-D series are the high-speed high-density devices fabricated in AMI's state-of-the-art 3-micron CMOS processes.

Table 1. 5-Micron Gate Array Family

Circuit	Equivalent Two-Input Gates	Pads	LS Output Drivers	TTL Output Drivers
UA-1	300	40	17	20
UA-2	400	46	23	20
UA-3	540	52	25	24
UA-4	770	62	31	28
UA-5	1000	70	35	32
UA-6	1260	78	39	36

### 5-Micron Gate Array Family

The family of 5-micron CMOS products is offered in six configurations, summarized in Table 1, with circuit complexities equivalent to 300, 400, 540, 770, 1000, and 1260 two-input gates, respectively. All pads can be individually configured as inputs, outputs, or I/O's. Input switching characteristics can be programmed for either CMOS or TTL compatibility. LS buffer output drivers will support CMOS levels of two low power schottky TTL loads. TTL buffer outputs will also provide CMOS levels and are capable of driving up to six LS TTL loads. All output drivers can be programmed for tri-state or open drain (open collector) operation as required.

The CMOS technology used for these products is AMI's state-of-the-art 5-micron, oxide-isolated, silicon gate CMOS process. This process offers all the conventional advantages of CMOS—i.e., very low power consumption, broad power supply voltage range (3V to 12V  $\pm 10\%$ ), and high noise immunity—as well as dense circuits with high performance. Gate propagation delays are in the five to ten ns range for 5 volt operation at room temperature. AMI gate array products can be supplied in versions intended for operation over the standard commercial temperature range ( $0^{\circ}\text{C}$  to  $+70^{\circ}\text{C}$ ), the industrial range ( $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ ), or the full military range ( $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ). MIL-STD-883 Class B screening, including internal visual inspection and



## Gate Arrays

high temperature burn-in, is offered. Similarly, customer-specified high reliability screening is available for commercial and industrial applications.

D.C. characteristics for the 5-micron gate array family are summarized in Table 2.

**Table 2**

Logic Element	2-Input Gate Equivalent
2-Input NOR	1
2-Input NAND	1
3-Input NOR	1.5
3-Input NAND	1.5
INVERTER	.5
D FLIP-FLOP W/RESET	5
D FLIP-FLOP W/SET-RESET	6
J-K FLIP-FLOP	8
CLOCKED LATCH	2.5
EXCLUSIVE OR	2.5
SCHMITT TRIGGER	2
4-BIT BCD CNTR W/RESET	27
TRANSMISSION GATE	.5

The current AMI array family, 300 gates to 1260 gates, is run in a 3-12V CMOS process (internally coded as CVA process). AMI is currently optimizing this array family for 3V to 5V operation (internally coded as CVH process). This new family, called UA-300 through UA-1260, is to be functionally identical to the existing UA-1 through UA-6. All design tools are interchangeable.

Customers who require 3V to 5V operation will be able to use the CVH family now. This optimized process will result in a 25-50% performance enhancement for 5V gate array designs. Customers who require operating voltages greater than 5V will continue to use the UA-1 to UA-6 CVA family.

In conjunction with these arrays, AMI has developed a set of "functional overlays." These are basic logic element building blocks—e.g. two input and larger gates of various types, flip-flops, and so forth—from which complete logic designs can be developed. Each functional overlay corresponds to a metal interconnect pattern that is superimposed on a set of uncommitted transistors (and polysilicon underpasses) in the array to implement the logic element. Typical functional overlay logic elements and the number of two-input gate equivalents they utilize are shown in Table 3.

Currently over 75 functional cells exist for this family.

### 3-Micron Gate Array Family

As part of AMI's long range semi-custom strategy in MOS/VLSI, AMI will continue to introduce new gate array

products. These new products will offer performance and cost advantages not currently realizable. In conjunction with these new array products, AMI has introduced computer-aided design tools to automate the entire gate array design process.

The newest gate array family is the high-performance GA and GA-D series which is based on AMI's 3-micron CMOS silicon gate process technology. With a 3-micron drawn geometry, it is equivalent to a 2-micron effective channel length which is the state-of-the-art.

The AMI GA and GA-D series are designed for 5V operation over military temperature range (-55 to 125°C). Besides high speed (2 to 3ns typical delay) and high density (up to 5K gates), it features total I/O flexibility in that each I/O pad can be one of any 13 options.

The single metal version provides up to 2500 gates and the double metal GA-D version 5000 gates. See Table 3 for configurations.

**Table 3. 3-Micron Gate Array Family**

Process	Product No.	Gates	Pads
Single Metal	GA-2500	2500	84
	GA-2000	2025	74
	GA-1500	1500	64
	GA-1000	1020	52
	GA-500	540	40
Double Metal	GA-5000D	4995	134
	GA-4000D	4012	120
	GA-3000D	3080	102
	GA-2000D	2016	84
	GA-1000D	1024	64

In conjunction with these new array products, AMI will have a complete powerful set of design automation software to allow users complete design flexibility. Using a terminal tied to a central AMI owned or customer owned minicomputer or mainframe, the user will have access to a complete set of design automation software tools including:

- Schematic digitization and capture
- Logic simulation
- Circuit simulation
- Test vector generation
- Interactive or autoplacement and route
- Auto continuity checking

These tools will allow the user to partially or fully automate the design task for maximum flexibility.

### Customer Interfaces

Unlike other gate-array vendors, AMI offers all the services required to develop and fabricate a semi-custom circuit. In-house capabilities include circuit and logic simulation, mask making, full wafer fabrication, assembly, and test.

Gate array developments are supported in two basic ways:

1) The **AMI-Tooled Interface** is available to customers with little or no experience in semi-custom circuit development, or to experienced users with limited in-house capabilities. In these cases AMI requires a logic diagram or breadboard schematic of the circuit to be integrated, along with D.C. and A.C. specifications, and functional patterns in I/O format.

At the end of the development phase, the customer receives 25 fully tested prototypes, assembled in ceramic packages, which have been subjected to our stringent Q.A. screening requirements.

2) The **Customer-Tooled Interface** is more applicable to customers with multiple circuit development requirements and suitable in-house design capabilities. For these customers AMI will supply the necessary software tools which have been written in FORTRAN and PASCAL languages so they will be compatible not only with PRIME computers, as is presently the case, but also with VAX and IBM minis. The user in turn provides AMI with either a database tape in CALMA format, or an ELECTROMASK or DAVID MANN compatible pattern generator tape, along with test requirements. A debugged SENTRY program is preferred. AMI will then fabricate and test 25 prototypes which should be representative of the production units.

### Software Support

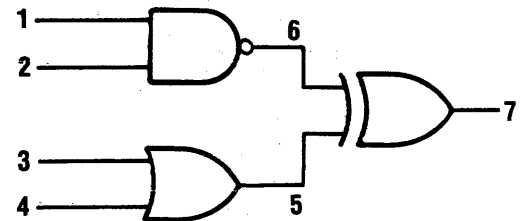
AMI CAD Technology (ACT™) is the most advanced integrated software package for CMOS/VLSI design available in the industry. It uses a common database for logic simulation, mask layout and test program generation. By operating from a common database, a gate array design can be converted into a standard cell or a fully crafted custom circuit with minimal risks.

The heart of the system is BOLT™ (Block Oriented Logic Translator) which is at the same time a hardware description language and a compiler. It allows the system designer to describe the logic network in a hierarchical fashion due to an unlimited macro nesting capability.

The common data base is created in a very simple fashion, consistent with the way that logic files are created in most timeshared logic simulators. Figure 2 shows a simple logic network and the corresponding BOLT™ file which allows programmability of rise and fall times for all gates.

GLIDE™ is the graphics editor on Prime used for interactive design of the single metal families of arrays. TRACE™ is a module that checks the layout database for continuity, and COMPARE™ verifies that the layout matches the BOLT™ logic file.

Figure 2.



6	°	NAND	1 2	(TDR = XX, TDF = XX);
5	°	OR	3 4	(TDR = XX, TDF = XX);
7	°	EXOR	5 6	(TDR = XX, TDF = XX);

GAPAR™ is the software module that allows automatic placement and routing of the double metal family of arrays. It is designed to complete at least 98% of the wiring connections on a 100% utilized array, based on a 90% confidence level using Donath-Heller-Mikhail two-dimensional model. The GAPAR™ system also offers interactive routing tools which can be used to complement the auto router when overflow conditions exist or to manually route critical delay paths.

CAPACITANCE™ calculates the actual capacitive load of the active gates from the layout database. The capacitance parameters are different for poly and metal interconnections, and for aluminum over diffusions or field oxide. In conjunction with DELAY™, it allows to accurately predict the dynamic behavior of the actual integrated circuit before it is fabricated.

TESTGEN™ is used to generate efficiently compressed functional test patterns based on the results of logic simulations from SIMAD™. This pattern compression minimizes the use of local memory on Sentry testers.

TESTPRO™ operates on Prime, allowing off-line generation of D.C. parametric tests in Factor language used in Fairchild test systems. Its output is merged with the compressed functional patterns from TESTGEN™, and the result is a full dynamic test program that can be tailored for use in any Sentry tester.

## Gate Arrays

### Packages

Pinout or lead count varies with die size and array complexity. The arrays are offered in standard plastic and ceramic dual-in-line packages with pin counts ranging from

16 to 64, JEDEC-Standard leadless and leaded chip carriers, miniflat packs to 84 pins, and pin grid arrays to 144 pins. AMI gate array products are also available in wafer or unpackaged die form.

### 5-Micron Gate Array Series

DC Characteristics—TTL Interface Specified @  $V_{DD} = +5V \pm 10\%$ ;  $V_{SS} = 0$ ; Temperature =  $-55^{\circ}C$  to  $+125^{\circ}C$

Symbol	Parameter	Min.	Typ.	Max.	Units
$V_{IH}$	Input High Voltage	2.0		$V_{DD}$	V
$V_{IL}$	Input Low Voltage	0.0		0.8	V
$V_{OH}$	Output High Voltage (LS Buffer $I_{OH} = -700\mu A$ ) (T Buffer $I_{OH} = -1.5mA$ )	2.7			V
		2.4			V
$V_{OL}$	Output Low Voltage (T Buffer $V_{OL} = 2.4mA$ ) (LS Buffer $I_{OL} = 0.8mA$ )			0.4	V
				0.4	V
$I_{OZ}$	3-State Output Leakage $V_O = 0$ or $V_{DD}$	-10	1	10	$\mu A$

### DC Characteristics—CMOS Interface

Sym.	Parameter	$V_{DD}$	Limits						Units	Condition	
			*T Low		25°C			*T High			
			Min	Max	Min	Typ	Max	Min			Max
$I_{DD}$	Quiescent Device Current	5V		0.1	.001	0.1		1	$\mu A/gate$	$V_{IN} = 0$ or $V_{DD}$	
		10V		0.2	.002	0.2		2			
$V_{OL}$	Low Level Output Voltage			0.05				0.05	V	$I_O = 1\mu A$	
$V_{OH}$	High Level Output Voltage	5V	4.95		4.95			4.95	V	$I_O = -1\mu A$	
		10V	9.95		9.95			9.95	V		
$V_{IL}$	Input Low Voltage	5V	0.0	1.5	0.0		1.5	0.0	1.5	V	
		10V	0.0	3.0	0.0		3.0	0.0	3.0	V	
$V_{IH}$	Input High Voltage	5V	3.5	5.0	3.5		5.0	3.5	5.0	V	
		10V	7.0	10.0	7.0		10.0	7.0	10.0	V	
$I_{OL}$	Output Low (Sink) Current T Buffer LS Buffer	5V	3.2		3.2	4.8		2.4		mA	$V_O = 0.4V$ $V_O = 0.5V$ $V_O = 0.4V$ $V_O = 0.5V$
		10V	6.0		6.0	9.0		4.0		mA	
		5V	1.0		1.0	1.6		0.8		mA	
		10V	1.8		1.8	3.1		1.0		mA	
$I_{OH}$	Output High (Source) Current T Buffer LS Buffer	5V		-600				-600		$\mu A$	$V_O = 4.6V$ $V_O = 9.5V$ $V_O = 4.6V$ $V_O = 9.5V$
		10V		-1120				-1120		$\mu A$	
		5V		-300				-300		$\mu A$	
		10V		-560				-560		$\mu A$	
$I_{IN}$	Input Leakage Current			1			1		$\mu A$	$V_{IN} = 0$ or $V_{DD}$	
$I_{OZ}$	3 State Output Leakage Current			$\pm 1$			$\pm 1$		$\mu A$	$V_O = 0$ or $V_{DD}$	
$C_I$	Input Capacitance					5			pF	Any Input	

\*Military temperature range is  $-55^{\circ}C$  to  $+125^{\circ}C$   
Industrial temperature range is  $-40^{\circ}C$  to  $+85^{\circ}C$   
Commercial temperature range is  $0^{\circ}C$  to  $+70^{\circ}C$

## Gate Arrays

### Absolute Maximum Ratings

Supply Voltage, $V_{DD}$	- .5V to +7V
Input Voltage, $V_{IN}$	- .5V to $V_{DD} + .5V$
D.C. Input Current, $I_I$	$\pm 10\text{mA}$
Storage Temperature, $T_{STG}$	-65° to +150°C

### 3-Micron Gate Array Series

D.C. Electrical Characteristics:  $V_{DD} = 5V \pm 10\%$ ,  $V_{SS} = 0V$ ,  $T_A = -55^\circ$  to  $+125^\circ\text{C}$

Symbol	Parameter	Min.	Typ.	Max.	Units	Conditions
$I_{DD}$	Quiescent Supply Current		10	50	$\mu\text{A}$	$V_I = V_{DD}$ or $V_{SS}$
$V_{OL}$	Low Level Output Voltage			.05	V	$I_{OL} = 1\mu\text{A}$
				.4	V	$I_{OL} = 3.2\text{mA}$
$V_{OH}$	High Level Output Voltage	4.95			V	$I_{OH} = -1\mu\text{A}$
		2.40			V	$I_{OH} = -5\text{mA}$
$V_{IL}$	Low Level Input Voltage	- .5		.8	V	TTL Interface
		- .5		1.5	V	CMOS Interface
$V_{IM}$	High Level Input Voltage	2.0		$V_{DD} + .5$	V	TTL Interface
		3.5		$V_{DD} + .5$	V	CMOS Interface
$I_{IN}$	Input Leakage Current			1	$\mu\text{A}$	$V_{IN} = V_{DD}$
$I_{OZ}$	High Impedance Output Leakage Current	-10	.001	10	$\mu\text{A}$	$V_{OH} = V_{DD}$ or $V_{SS}$
$C_{IN}$	Input Capacitance		5		pF	Any Input

**Table 4. Propagation Delay Characteristics for AMI Gate Arrays**

A.C. Characteristics:  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = 4.5V$ , F.O. = 2

Function		Gate Delays			Units
		5 $\mu$	3 $\mu$ S.M.	3 $\mu$ D.M.	
INVERTER	L-H	6.0	4.5	3.4	ns
	H-L	5.3	1.3	1.0	ns
2INPUT NAND	L-H	9.0	5.8	4.4	ns
	H-L	9.0	1.8	1.4	ns
2INPUT NOR	L-H	12.0	8.4	6.3	ns
	H-L	8.2	1.5	1.1	ns
<b>D-TYPE F/F</b>					
MAX FREQUENCY		15	36	45	MHz
TTL INPUT BUFFER		18	7	5	ns
<b>OUTPUT BUFFER</b>					
DELAY ( $C_L = 30\text{pF}$ )		20	9	6.7	ns

**NOTE:** BASED ON WORST CASE PROCESS PARAMETERS. FOR TYPICAL DATA, DIVIDE GATE DELAYS BY 1.8.



## Standard Cells

### AMI's Standard Cells

#### Program Description

The cells in the Standard Cell Program are basic logic elements such as gates, flip-flops, register counter bits and I/O devices. Each cell has been previously designed and analyzed for performance. Complex digital functions can be rapidly implemented by interconnecting the various cells. Most cells have a series 4000 CMOS equivalent and a 74LS TTL equivalent for ease of breadboarding.

The cells are initially designed on an interactive color graphics terminal — AMI's SIDS (symbolic interactive design) system. The cell library is maintained within the SIDS data base. If other CAD (Computer-Aided Design) systems are used internally by a customer, the cell library can be digitized onto these systems.

Using the SIDS system, new cells or modifications of existing cells can be generated and added to the cell library rapidly, without any hand layout. Similarly, non-cell functions such as analog elements or memory arrays, RAM or ROM, can be designed using SIDS and merged with standard cells.

This ability to add special features efficiently makes AMI's standard cell program far more flexible than other manufacturer's cell development systems.

Table 5. Standard Cell Offerings

Part No.	Process Technology		Availability
XXXX-001	5μ CMOS	Single Metal	NOW
XXXX-001	4μ NMOS	Single Metal	NOW
XXXX-001	3μ CMOS	Single Metal	NOW
XXXX-001	3μ CMOS	Double Metal	NOW
XXXX-001	2μ CMOS	Double Metal	2Q'84

#### Performance

AMI's standard cell circuits can be successfully used in digital circuits with operating speeds up to 10MHz for NMOS (or CMOS at 10V V<sub>CC</sub>) up to 40MHz for CMOS (5V V<sub>CC</sub>). It should be emphasized that if only a small portion of the circuit requires faster performance, this portion can be "customized" either by creating special cells or by designing circuitry outside of the cell structure. AMI will review customer logic without obligation to determine if a cell design is feasible.

#### Features

The high speed, high performance 3-micron cell families are also available. These cells are designed in AMI's 3-micron single and double layer metal process. They have been characterized for 3-5V operation and over the full military temperature range. As in 4000 series CMOS, power for static operation is near zero.

#### 3-Micron Single Metal Standard Cells

- State-of-the Art CMOS Technology
- 3-Micron Single Metal Cell as Basic Building Block
- Same Macros and Performance for Cells and Arrays

#### 3-Micron Double Metal Standard Cells

- Automatic Placement and Routing
- Cells are Designed to Exceed the Performance of 3-Micron DLM Gate Arrays
- Fixed Cell Height/Variable Width
- Characterized at 25°C, 70°C, and 125°C with V<sub>DD</sub> = 4.5V

#### NMOS Cells

This family is AMI's most complete cell family with over 135 cells in 4 speed/power combinations. V<sub>DD</sub> and V<sub>SS</sub> connections are metal lines running horizontally across the cell. Two other metal lines, usually used for clock signals, also run across the cell. Typical and worst case inverter delays are 5.5ns and 10.0ns, respectively.

The NMOS cells are implemented using a 4-micron silicon gate process and can be used over the full military temperature range (-55°C to 125°C). Operating voltage is 5V ± 10%. The NMOS cells have been designed with three power/speed options. The fastest cells also have the highest power consumption and use the most area. Therefore, the fast cells should be used only where circuit performance requires high speed. Most circuits are optimized by using a combination of low power, standard and high speed cells.

#### CMOS Cells

The CMOS cells are designed using a 5-micron silicon gate oxide-isolated process and using 3-micron single and double layer metal process. These processes are well suited for analog circuitry and some analog cells will be added to the digital cells. The 5-micron CMOS cells are characterized for 3V to 12V operation over the full military temperature range (-55°C to 125°C). CMOS cells are generally used where low power battery operation or backup are required.

#### AMI Standard Development

The circuit user provides a completed logic diagram and a circuit specification. AMI performs all other design activity including MOS logic design, circuit design, layout, mask generation and fabrication of wafers. This development option is recommended for most users desiring to build a single LSI device and for multiple circuit users who do not wish to become directly involved in the MOS circuit development.



## Standard Cells

### Developing Your Standard Cell Design

AMI offers three basic options for developing a standard cell custom circuit. All of these apply to the Development Flow Diagram of Table 7.

Table 7. Gate Array/Standard Cell Development Flow

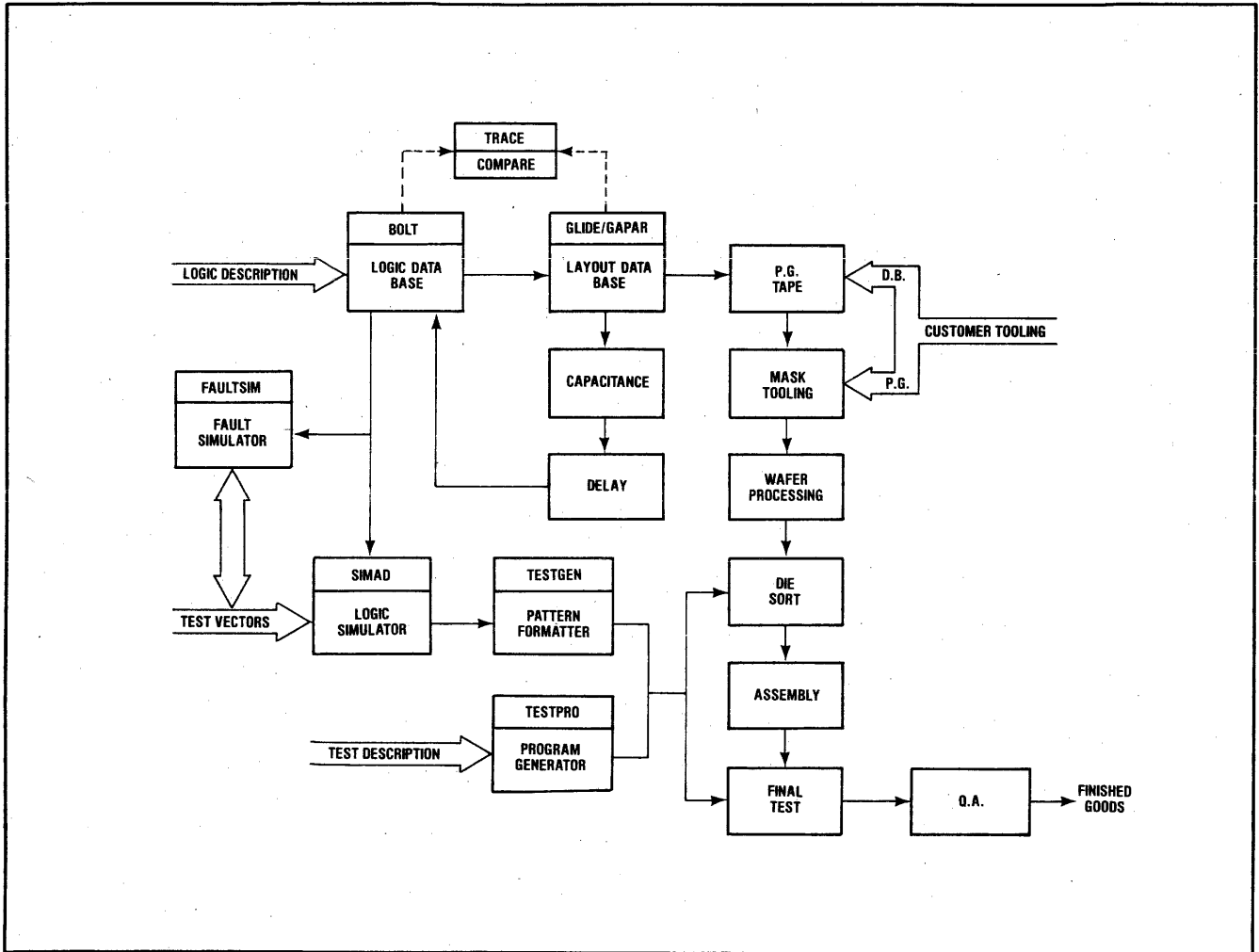


Table 8. AMI Standard Cell  
3-Micron Single Metal Interface

No. of Gates	Logic Diagram Input		Net List Input	
	Cost	Max. Span	Cost	Max. Span
500	\$19.7K	9 Weeks	\$15.7K	7 Weeks
1000	\$22.8K	11 Weeks	\$16.8K	9 Weeks
2000	\$31.3K	13 Weeks	\$21.3K	11 Weeks



# We're the last people you need to call

---

**L**ook around; compare price, compare yield, compare turnaround time . . . then call BRI. We welcome your toughest VLSI design problem. Gate arrays, standard cell, PLS/PAL, Bi-CMOS, you name it. We've got a fully computerized design solution and a fabrication technique that will give you what you need, when you need it, at the right price. Volume is no obstacle.

**BRI** has combined a fully computerized and world-wide-accessible VLSI design system with high-volume, high-yield fabrication facilities to provide our clients with products offering both reduced die size and superior electrical characteristics. From initial design through final component assembly, or at any step in the process, BRI can bring an imaginative and cost-effective solution to your VLSI requirements. We don't have to be the first people you call but, when it comes to a final decision, we're determined to be the last.

# for Semi-Custom I.C.s

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## ◆ Standard Cell

BRI's computer aided VLSI design features an extensive cell library which includes 7400 series, ROM/RAM and analog functions. This allows customized circuit design at lower cost and in less time. The 2 micron HCMOS dual layer metal process technology and advanced design techniques provide both increased circuit performance and reduced die size.

## ◆ Gate Array

The 2 micron HCMOS dual layer metal process technology is supported by a fully automated design system to provide a family of gate arrays with 1.6 ns maximum propagation delay. All BRI gate array products are specifically designed for quick turn-around, high performance and low cost.

## ◆ Analog/Digital Gate Array

Fabricated with the Bi-CMOS process technology, these arrays offer analog functions with true bipolar characteristics and low power CMOS digital functions in one cost-effective device. This allows the customer to maximize system integration while protecting his proprietary design.

## ◆ PAL/PLS

These 2 micron HCMOS mask programmable products provide bipolar speed and CMOS power dissipation for greater reliability and lower cost. They are pin compatible with Signetics PLS and MMI PAL series.

For full facts on these and other BRI custom and semi-custom products, give us a call.

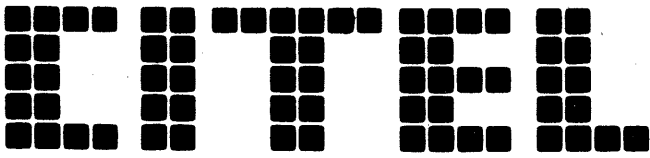
# BRI



**BARVON RESEARCH, INC.**

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# MA Series High Speed CMOS Gate Arrays 3 $\mu$ Silicon Gate CMOS

Preliminary

## Benefits

- Integrate HCMOS or standard TTL functions easily
- Reduce component count
- Reduce power dissipation and consumption
- Introduce new products quickly
- Improve system reliability
- Protect innovative designs longer

## Features

- Typical propagation delay ( $t_p$ ) < 2.0 nsec
- Circuit/package flexibility
- Quiescent current: 0.01  $\mu$ A/gate
- Dedicated VLSI CAD system with complete software support
- Library of pre-designed macro cells
- Input protection on all pads
- Rapid design cycle—from 6 to 12 weeks depending on array size

## General Description

The MA series of gate arrays is produced on Citel's 3 $\mu$  silicon gate CMOS process, a state-of-the-art technology featuring oxide isolation, isoplanar structures and shallow functions for high speed and high density. As a fully licensed second source for both design and complete fabrication of Matra Harris gate arrays, Citel offers an extensive library of fully characterized macro cells which enables a rapid, yet reliable, logic implementation of any cell on any array.

Gate array design activity is performed on the Gate-mark VLSI design system which includes such capabilities as schematic or net list entry, logic simulation, interactive layout, automatic place and route and a layout/net list check.

In addition to design and wafer fabrication, Citel offers a wide range of package options and a complete S-VII test capability for both wafer sort and final test. The I/O pad-to-gate ratio permits wide flexibility in package selection (e.g., the MA0400 fits a standard 20-pin DIP yet also provides 40 pads for those with high I/O needs). The MA Series currently includes four arrays ranging from 200 to 1200 gates. Additional semi-custom arrays and a complete library of standard cells for full custom CMOS design are in development.

## Citel 3 $\mu$ CMOS Gate Array Family (MA Series)

Part No.	Gate Count	Pad Count	Packages			Cells Available
			Plastic DIP	Ceramic	LCC	
MA0200	228 & 28 I/O	32	8-28 pins	14-28 pins	14-28 pins	AND/OR/INV — 14 NAND/NOR/AND-OR-INV/EXOR — 19 Latches — 5
MA0400	380 & 36 I/O	40	8-40 pins	18-40 pins	28-40 pins	JK/D/RS Flip Flops — 17 MSI Functions — 10 (Counters, Adders, Registers)
MA0800	754 & 50 I/O	54	22-40 pins	22-48 pins	32-64 pins	I/O Cells — 10 Misc. Functions — 7
MA1200	1139 & 62 I/O	66	24-40 pins	24-64 pins	40-64 pins	



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**Absolute Maximum Ratings**

Supply Voltage .....	-0.5 to +7.0V
Input/Output Voltage .....	-0.3 to $V_{DD} + 0.3V$
Storage Temperature .....	-65°C to +150°C

**DC Electrical Characteristics** ( $V_{DD} = 5V \pm 10\%$ ; -55°C to +125°C)

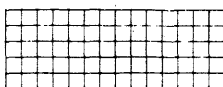
Parameter	Symbol	Min.	Typ	Max	Units	Conditions
Low-level Input Voltage (TTL)	$V_{IL}$	—	—	0.8	V	Input Buffer Cells
Low-level Input Voltage (CMOS)	$V_{IL}$	—	—	1.5	V	$V_O = 4.5V; I_O = 0$
High-level Input Voltage (TTL)	$V_{IH}$	2.2	—	—	V	Input Buffer Cells
High-level Input Voltage (CMOS)	$V_{IH}$	3.5	—	—	V	$V_O = 0.5V; I_O = 0$
Low-level Output Voltage	$V_{OL}$	—	0.2	0.4	V	$I_{OL} = 3.2 \text{ mA}$
High-level Output Voltage	$V_{OH}$	3.9	—	—	V	$I_{OH} = -100\mu A$
Input Leakage Current	$I_{IL}$	-1	0.01	+1	$\mu A$	$V_{IN} = 0V \text{ or } V_{DD}$
Standby Supply Current	$I_{DD}$	—	0.01	—	$\mu A/\text{Gate}$	

**AC Electrical Characteristics** ( $V_{DD} = 5V$ ; Fan-out = 2 gates;  $T_{PD} = \frac{1}{2}(T_R + T_F)$ )

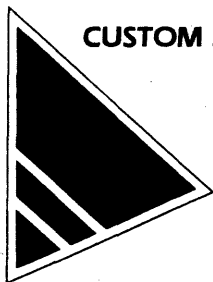
Parameter	Typ (25°C)	Max (0° to 70°C)	Max (-55° to 125°C)	Units
Propagation Delay ( $t_p$ )				
Inverter	1.3	3.0	4.0	nsec
4-Input NAND	1.9	4.6	6.0	nsec
4-Input NOR	4.0	10.0	11.0	nsec
Input Buffer	2.2	7.0	9.0	nsec
Output Buffer (50pF)	13.0	20.0	25.0	nsec
Dynamic Supply Current	5.0 (max)	—	—	$\mu A/\text{MHz}/\text{Gate}$

Citel's capabilities also include full-custom MOS and bipolar integrated circuits and Universal Programmers. Complete information available upon request.

This data sheet is issued to provide information only. Citel reserves the right to alter without notice the specifications, design, price or condition of supply of this product.  
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# CMA Silicon Gate HCMOS Gate Arrays

## CMA 2000 Series

### GENERAL

The CMA 2000 Series of gate arrays from Custom MOS Arrays (CMA) are a high-performance family of gate arrays, with complexity ranging from 400 to 2400 gates and pin count ranging from 40 to 110 pins.

The CMA 2000 Series uses 3-micron technology with single-level metal and programmable via holes for interconnects, and offers Schottky TTL speeds and CMOS power consumption.

### FEATURES

- 400-2400 Gate Complexity
- Up to 110 Input/Output (I/O) Pins
- Schottky TTL Speeds
- Reliable and Proven 3-micron HCMOS Technology
- Fully Characterized to Military Temperature Range
- Hermetic and Plastic Packaging
- Supports 7400/4000 SSI/MSI Functions
- Full TTL/CMOS Interface Compatibility
- Ultra-Low Leakage and Standby Currents
- Proprietary Latch-Up (SCR) Resistant Design
- Proprietary Static Protection Circuits
- Hermetic Packaged Parts are Processed to MIL-STD-883, Class C

### PRODUCT DESCRIPTION

The CMA 2000 Series of gate arrays are manufactured using proven oxide-isolated HCMOS technology. Single-level metal interconnect enables CMA to obtain higher yields and hence lower production cost.

The gate arrays are organized as arrays of cells, and individual N and P channel transistors are configured into a variety of logic elements called macrocells. CMA's current macrocell library is detailed in Appendix 1.

A variety of Input/Output (I/O) configurations is available:

- All I/O pads are configurable as inputs, outputs or bi-directional tri-states.
- All I/O pads can have active pull-ups or pull-downs.
- Any input may make use of TTL input shifters for TTL requirements.
- All I/O pads can drive multi-TTL loads and may be paralleled for increased drive.

The CMA 2000 Series of gate arrays are ideal in applications where speed, space and power are at a premium.

### ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Conditions	Limits	Unit
V <sub>CC</sub>	Supply voltage	With respect to GND	-0.3~7	V
V <sub>I</sub>	Input voltage		-0.3~7	V
T <sub>opg</sub>	Operating ambient temperature (Military)		-55 ~ +125	°C
T <sub>opg</sub>	Operating ambient temperature (Commercial)		0 ~ 70	°C
T <sub>stg</sub>	Storage temperature (Hermetic)		-65 ~ 150	°C
T <sub>stg</sub>	Storage temperature (Plastic)		-40 ~ 125	°C

### PRODUCT OUTLINE

Array Number	Complexity	Max. I/O Pads	Technology	Gate Speed*
CMA 2004	400	40	3-micron	4.5 ns
CMA 2008	800	60	3-micron	4.5 ns
CMA 2011	1100	68	3-micron	4.5 ns
CMA 2015	1500	80	3-micron	4.5 ns
CMA 2024	2400	110	3-micron	4.5 ns

\*Notes on Gate Speed: 1. Two-input NAND gate, fanout = 3 with 3 mm interconnect. 2. T<sub>A</sub> = 70°C, V<sub>DD</sub> = 5V ± 10%. 3. T<sub>pd</sub> = ½ \* (T<sub>pLH</sub> + T<sub>pHL</sub>).

## PROPAGATION DELAYS

In CMOS technology the propagation delay is a function of:

- Fanout
- Interconnect Routing
- Supply Voltage
- Temperature
- Processing
- Input Slew Rate/Polarity

CMA's CAD system takes all of these variables into account (except Input Slew Rate/Polarity) during simulation.

After the design netlist has been entered into CMA's CAD system the worst-case delay of each cell is automatically determined, taking into account the affects of fanout and estimated interconnect capacitance.

After interconnect routing is completed, the actual interconnect loading is calculated automatically from the database and automatically inputted into the simulator and final design verification is completed.

The effects of supply voltage and temperature are illustrated in Figures 1 and 2 below. A rule of thumb for processing variation is 35% to give worst-case delay figures.

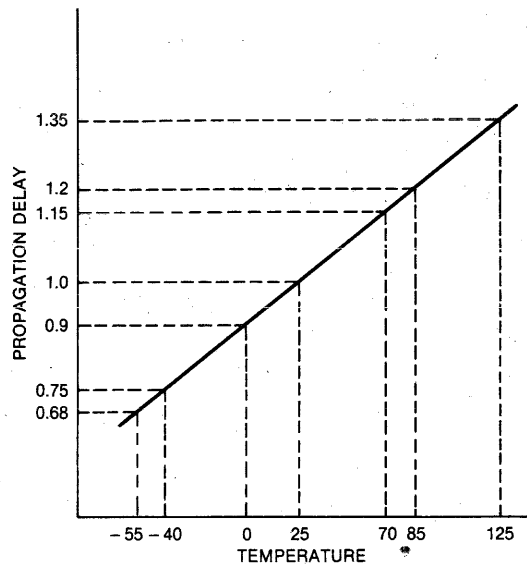


Figure 1. CMOS Delay vs. Temperature

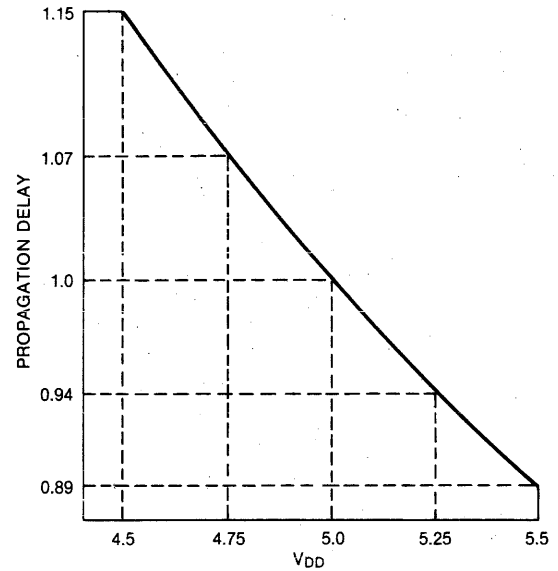


Figure 2. CMOS Delay vs. Supply Voltage

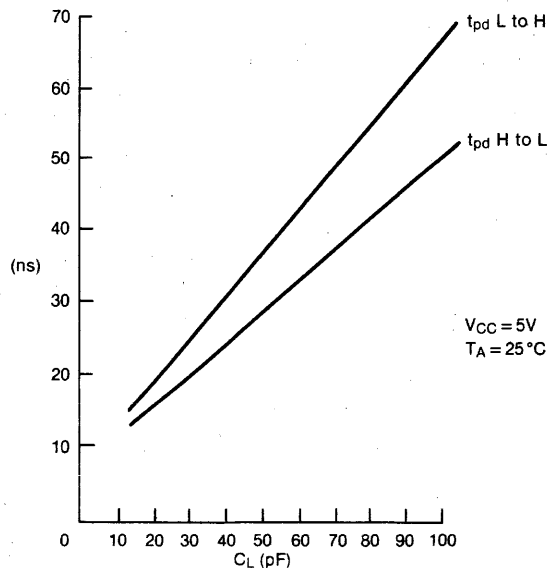


Figure 3. CMOS Driver Delay vs. Output Loading

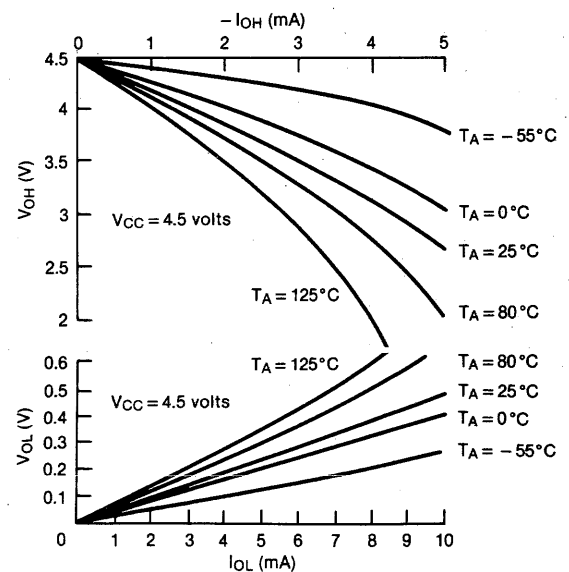


Figure 4. Output Current Drive

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## DC CHARACTERISTICS (T<sub>A</sub>=0~70°C, V<sub>CC</sub>=5V ± 10%)

Symbol	Parameter	Test Conditions	Limits			Unit
			Min.	Typ.	Max.	
V <sub>IH</sub>	High-level input voltage		2.0		V <sub>CC</sub>	V
V <sub>IL</sub>	Low-level input voltage		-0.3		0.8	V
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = +1 mA	2.4			V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 4 mA			0.4	V
I <sub>IL</sub>	Input current	V <sub>I</sub> = 0~5.5V			±10	μA

## POWER CONSUMPTION

In CMOS, power dissipation is a function of three elements:

- Charging and discharging circuit nodes, usually 90% total power.
- Overlap currents when P and N transistors switch, usually 10% power.
- Leakage (DC power). This is negligible, a few micro-watts.

Charging and discharging circuit nodes and the associated power consumption may be calculated using the equation  $P = CV^2f$ , where C is node capacitance, V is supply voltage, and f the operative frequency. The following example illustrates a typical power supply calculation.

Parameter	Gate Array	
	CMA 2004	CMA 2024
No. of Gates	400	2400
Percent Utilization (80%)	320	1920
No. of Switching Nodes (15%)	48	288
No. of Outputs Utilized	20	80
Output Load Capacitance	45 pF	45 pF
No. of Outputs Switching (20%)	4	16
Supply Voltage	5V	5V
Core Dissipation per MHz	0.6 mW	3.6 mW
Output Dissipation per MHz	4.5 mW	18 mW
At 10 MHz Clock Speed	51 mW	216 mW
At 30 MHz Clock Speed	153 mW	648 mW

Note: Overlap power is taken into account by increasing the nodal capacitance in the above calculation.

## PRODUCT OPTIONS

The CMA 2000 Series are offered in a variety of operating temperature ranges and packages.

- Military Operating Range: -55°C to +125°C
- Industrial Temperature Range: -40°C to +85°C
- Commercial Temperature Range: 0°C to +70°C
- MIL-STD-883 Levels B and C Screening
- Most commercial packages are available as well as unpackaged die.

## EQUIVALENT GATE COMPLEXITY

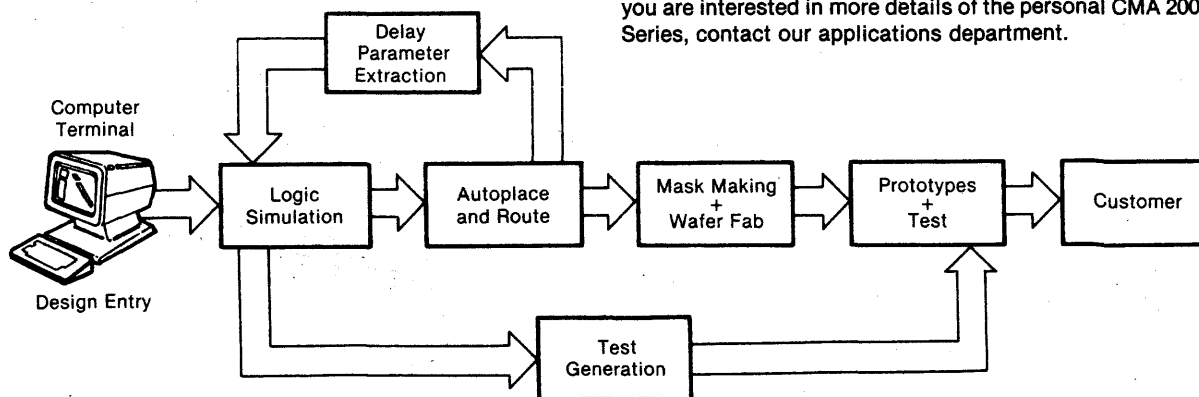
One of the first tasks in converting your logic functions into a gate array is to estimate its logic complexity. CMA defines gate count in terms of 2-input NAND gate equivalents.

Tables 1 and 2 list the approximate equivalent CMOS 2-input NAND gate count of most of the popular 7400 and 4000 Series SSI/MSI functions.

## GATE ARRAY DEVELOPMENT SYSTEM CMA 2000 SERIES

The CMA CAD system is configured to allow generation of error-free designs. The desired logic functions are entered into the system and logical and timing verification is completed via CMA's simulator. The simulated input patterns and corresponding outputs are checked for completeness and other tests added until the final test patterns are completed. The flow diagram below illustrates the design cycle.

CMA's CAD team is currently working on a personal gate array development system for customer design verification. If you are interested in more details of the personal CMA 2000 Series, contact our applications department.



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**Table 1. 7400 Equivalent Gate Count**

Part No.	Gate Equivalent	Part No.	Gate Equivalent	Part No.	Gate Equivalent	Part No.	Gate Equivalent
7400	4	7470	16	74152	24	74253	40
7401	4	7471	15	74153	27	74257	36
7402	4	7472	14	74154	33	74258	34
7403	4	7473	20	74155	21	74259	48
7404	4	7474	15	74156	21	74260	6
7405	3	7475	14	74157	22	74261	58
7406	3	7476	20	74158	23	74265	7
7407	6	7477	14	74159	18	74266	10
7408	6	7478	20	74160	68	74273	50
7409	6	7480	17	74161	68	74274	70
7410	6	7481	80	74162	68	74275	70
7411	8	7482	27	74163	68	74276	40
7412	6	7483	57	74164	43	74278	34
7413	14	7484	80	74165	85	74279	10
7414	30	7485	73	74166	62	74280	24
7415	8	7486	12	74167	52	74281	170
7416	3	7487	19	74168	80	74283	59
7417	6	7489	320	74169	80	74284	58
7420	6	7490	40	74170	116	74285	58
7421	5	7491	43	74172	205	74289	280
7422	4	7492	41	74173	56	74290	48
7423	6	7493	41	74174	37	74293	45
7425	6	7494	42	74175	25	74295	57
7426	4	7495	42	74176	42	74298	32
7427	6	7496	44	74177	41	74299	188
7428	4	7497	120	74178	56	74323	188
7430	6	7498	33	74179	56	74348	64
7432	4	7499	46	74180	20	74351	124
7433	4	74100	18	74181	100	74352	26
7437	4	74102	14	74182	30	74353	39
7438	4	74103	20	74183	16	74362	44
7440	4	74106	20	74184	100	74363	28
7442	29	74107	20	74185	100	74364	48
7443	29	74108	20	74190	80	74365	37
7444	29	74109	21	74191	76	74366	40
7445	29	74110	14	74192	62	74367	37
7446	45	74111	20	74193	58	74368	40
7447	45	74116	26	74194	70	74373	58
7448	45	74120	15	74195	46	74374	74
7449	45	74135	24	74196	42	74375	16
7450	6	74136	12	74197	41	74376	40
7451	5	74138	23	74198	92	74377	48
7452	8	74139	20	74199	85	74378	34
7453	7	74141	16	74225	450	74379	24
7454	7	74142	90	74226	116	74381	150
7455	6	74143	45	74245	83	74386	12
7460	6	74144	98	74246	51	74390	86
7461	6	74145	24	74247	51	74393	80
7462	9	74147	35	74248	51	74395	50
7464	8	74148	35	74249	51	74398	22
7465	8	74150	60	74251	37	74399	22
		74151	24			74670	144

**Table 2. 4000B Equivalent Gate Count**

Part No.	Gate Equivalent	Part No.	Gate Equivalent	Part No.	Gate Equivalent	Part No.	Gate Equivalent
4000	5	4038	74	4153	32	4515	69
4001	4	4040	74	4155	22	4516	66
4002	4	4041	4	4157	24	4518	26
4006	108	4042	12	4158	24	4519	28
4007	2	4043	20	4160	64	4520	54
4008	32	4044	20	4162	64	4521	123
4009	5	4051	6	4163	64	4522	44
4010	5	4068	4	4164	51	4526	47
4011	4	4069	3	4166	84	4530	16
4012	4	4070	12	4174	38	4531	36
4013	10	4071	6	4175	25	4532	39
4014	43	4072	4	4192	60	4534	190
4015	62	4073	8	4193	56	4539	24
4017	45	4074	14	4273	50	4554	40
4018	52	4075	8	4373	58	4555	22
4019	12	4076	41	4374	74	4556	22
4020	40	4077	12	4375	14	4558	65
4021	43	4078	4	4386	12	4560	78
4022	35	4081	4	4501	6	4580	187
4024	44	4082	4	4504	33	4581	100
4026	65	4089	84	4505	271	4582	31
4027	20	4093	32	4506	10	4583	23
4028	22	4094	86	4508	24	4584	30
4029	68	4099	64	4510	94	4585	35
4030	12	4106	30	4511	136	4597	103
4032	68	4107	18	4512	31	4598	74
4034	110	4138	20	4514	69	4599	42
4035	48	4139	10				

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Fairchild's high performance gate arrays feature a wide range of gate counts and offer minimal circuit delays. The FGT0750 and the FGE1000 and FGC6000 series complement the FAST™, F100K ECL and HCMOS logic families respectively, allowing for easy integration

of standard SSI/MSI functions into a unique LSI chip. Designs for all arrays are implemented using pre-defined SSI/MSI functions (macros). Fairchild's complete Computer-Aided Design System, FAIRCAD™, reduces development time and assures first time success.

Product	Gate Equiv.	Technology	I/O Levels	Typical Internal Gate Delay (ns)	Typical Buffer Delay		P <sub>D</sub> Typ (W)†	I/O	Packaging	Available
					Input (ns)	Ouput (ns)				
FGE0020	20	ECL	100K ECL	0.35	N/A	0.55	0.5	21	24 F/P	Now
FGE0500	500	ECL	100K/10K ECL	0.3-1.5	N/A	0.55	1.5-4.0	72	84-Pin CC 84-Pin Grid	Now
FGE2000	2000	ECL	100K/10K ECL	0.3-1.5	N/A	0.55	4.0-9.0	126	132-Pin CC 132-Pin Grid	Now
FGT0750	1000	FAST	TTL	1.8	1.8	3.0	1.0	76	68-, 84-Pin CC 68-, 84-Pin Grid	Now
FGC0500	540	2μ CMOS	TTL/ CMOS	1.5	2.0	6.0		40	28, 48 DIP 44 F/P	4Q '83
FGC1200	1242	2μ CMOS	TTL/ CMOS	1.5	2.0	6.0		73	40, 68 DIP 68, 84 PGA 68, 84 LCC	3Q '84
FGC2400	2625	2μ CMOS	TTL/ CMOS	1.5	2.0	6.0		109	84, 120 PGA 68, 84 LCC 84 LDCC 64 DIP	1Q '84
FGC4000	4020	2μ CMOS	TTL/ CMOS	1.5	2.0	6.0		133	120, 144 PGA 144 LDCC 64 DIP	3Q '84
FGC6000	6000	2μ CMOS	TTL/ CMOS	1.5	2.0	6.0		161	172 PGA 172 LDCC 64 DIP	2Q '84

†For the CMOS arrays, maximum power dissipation = 20 μW/switching gate/MHz.

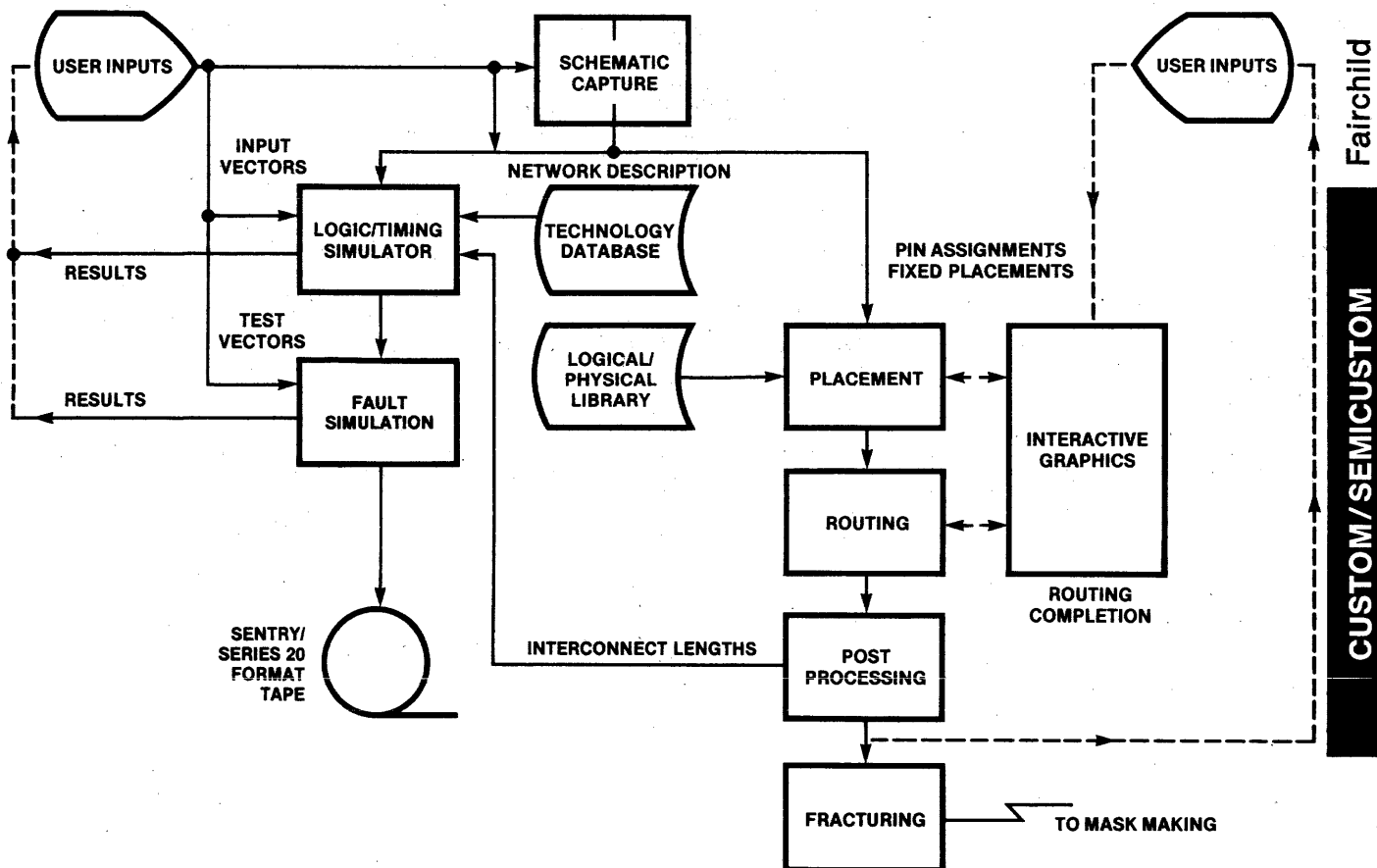
For further information on product availability, CAD training and design schedules write to: Gate Array Marketing, 1801 McCarthy Blvd., Milpitas, CA 95035, or call (408) 942-2672.

**FAIRCAD Computer-Aided Design System**

A user friendly computer-aided design system is employed in all phases of the design cycle to assure fast turnaround times. Major functions performed by the FAIRCAD system include schematic capture, logic and timing simulation, test vector grading, automatic placement and routing and design rules checking. The user's logic description is input either interactively using a graphics workstation or entered in netlist form. Fairchild's FAIRLOG<sup>SM</sup> simulator is used to verify array logic and timing. A fault simulator grades the effectiveness of a user-supplied test vector sequence. Placement and routing can be highly interactive processes, allowing the user to assign pins and pre-place critical path functions. Automatic routing

eliminates tedious manual steps, thereby reducing routing design time from weeks to minutes. Upon routing completion, the circuit is resimulated using actual macro interconnect delays as a final verification of functionality and system performance. Design-rules checking insures that process-imposed geometry constraints are not violated.

The FAIRCAD system uses a multiple CPU configuration to insure high system availability. Remote access via telephone link is available and additional design centers are being planned worldwide. The FAIRCAD system is product and technology transparent, allowing new gate arrays and technologies to be easily integrated into the system.

**FAIRCAD System**

# ABBREVIATIONS OF COMPANY NAMES

**Action Ins** Action Instruments  
**AD** Analog Devices  
**ADT** Advanced Digital Technology  
**Advent** Advent Products, Inc.  
**Alphatron** Alphatron  
**AMA** American Automation  
**AMD** Advanced Micro Devices  
**AMI** American Microsystems, Inc.  
**Amperex** Amperex Electronic Corp.  
**Analogic** Analogic  
**Analog Sys** Analog Systems  
**APC** Applied Micro Circuits  
**Apex** Apex Microtechnology  
**APM** Applied Microsystems Corp.  
**Appl Sys** Applied Systems Corp.  
**APT** Applied Microtechnology  
**Aptek** Aptek Microsystems  
**Array Tech** Array Technology  
**AWI** AWI Electronics

**Barvon** Barvon Research  
**Bedford** Bedford Computer Systems Inc.  
**Burr-Brown** Burr-Brown

**CAE** Computer Aided Engineering  
**Cal Devices** California Devices  
**Cermetek** Cermetek  
**CGRS** CGRS Microtech Inc.  
**Cherry** Cherry Semiconductor  
**CIC** Custom Integrated Circuits  
**CirTech** Circuit Technology  
**Citel** Citel  
**Comlinear** Comlinear Corporation  
**CMA** Custom MOS Arrays  
**Comark** Comark Corp.  
**Comdial** Comdial Semiconductor  
**Comp Auto** Computer Automation  
**Compas** Compas Microsystems  
**Cont Logic** Control Logic Inc.  
**Control Sys** Control Systems Microsystems Div.  
**CreMicro** Creative Micro Systems  
**Cromemco** Cromemco, Inc.  
**Cubit** Cubit Inc.  
**Curtis** Curtis Electro Devices, Inc.  
**Cybernetic** Cybernetic Micro Systems  
**Cybersys** Cybersystems  
**Cybertek** Cybertek Inc.

**Data General** Data General  
**Data I/O** Data I/O  
**Data Trans** Data Translation  
**Datel** Datel  
**Datricon** Datricon Corporation  
**DDC** Data Devices Corporation  
**DEC** Digital Equipment Corporation  
**Die-Tech** Die-Tech  
**Digelec** Digelec Corp.  
**Digitel** Digitel, Inc.  
**Dionics** Dionics Inc.  
**Dist Comp** Distributed Computer Systems  
**Divers Tech** Diversified Technology

**E-HI** E-H International, Inc.  
**EDI** Electronic Designs Inc.  
**Elind** Elind Elettronica Industriale  
**EMM-SESCO** EEM-SESCO  
**Emulogic** Emulogic Inc.  
**ETI Micro** ETI Micro  
**Exar** Exar Integrated Systems  
**Exel** Exel Microelectronics

**Fairchild** Fairchild  
**Ferranti** Ferranti Electric  
**Force** Force Computers  
**Fujitsu A** Fujitsu America  
**Fujitsu** Fujitsu Microelectronics, Inc.

**GI** General Instrument  
**GTE Micro** GTE Microcircuits

**Harris** Harris Semiconductor  
**Heurikon** Heurikon Corp.  
**Hilevel** Hilevel Technology, Inc.  
**Hitachi** Hitachi America, Ltd.  
**Holt** Holt Inc.  
**HP** Hewlett-Packard  
**Hughes** Hughes Aircraft, Solid State Products

**Hybrid Sys** Hybrid Systems  
**HyComp** HyComp

**ICC** International Cybernetics  
**IDT** Integrated Device Technology  
**IMI** International Microcircuits, Inc.  
**IMP** International

**IMS** International MicroSystems Inc.  
**Infosphere** Infosphere  
**Inmos** Inmos  
**IntCirEng** Integrated Circuit Engineering  
**IntCirSys** Integrated Circuit Systems  
**IntCompSys** Integrated Computer Systems  
**Int Tech** Integrated Technology Corp.  
**Intech** Intech Microcircuits Div.  
**Intel** Intel  
**Interdesign** Interdesign  
**Intersil** Intersil  
**Intronics** Intronics  
**ITT** ITT Semiconductors

**Kinetic Sys** Kinetic Systems  
**Kontron** Kontron Electronics

**Lambda** Lambda Semiconductor  
**Linear Tech** Linear Technology  
**LSI Comp** LSI Computer Systems  
**LSI Logic** LSI Logic Corporation

**Master Logic** Master Logic Corporation  
**Matrix** Matrix Corp.  
**Matrox** Matrox Electronic Systems  
**Maxim** Maxim Integrated Products  
**MCC** Micro-Computer Control  
**MCE** MCE Electronics  
**Micrel** Micrel  
**Micro Innov** Micro Innovators  
**Micropac** Micropac Industries  
**Micro Net** Micro Networks  
**Micro Pwr** Micro Power Systems  
**Micro Sci** Micro Sciences Corp.  
**Micro Tech** Microcircuits Technology  
**Micro-Link** Micro-Link Corporation  
**Micron** Micron Technology  
**MillerTron** MillerTronics  
**Miller** Miller Technology  
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**Mitsubishi** Mitsubishi Electronics  
**MNI** Monolithic Memories, Inc.

**Mostek** Monolithic Systems Corp.  
**Motorola** Mostek  
**MRC** Motorola Semiconductor  
**Murray** MRC Systems  
**Monosil** Murray Consulting

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**NCM** NCM Corp.  
**NCR** NCR Corp., Microelectronics Division  
**NEC** NEC Electronics  
**Nitron** Nitron

**OAE** Oliver Advanced Engineering  
**Octagon** Octagon Systems Corp.  
**OEI** Optical Electronics Inc.  
**Ohio Sci** Ohio Scientific  
**OKI** OKI Semiconductor  
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**Onset** Onset Computer Corp.

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## TTL Gate Arrays

# FUJITSU

### ■ B240, B350, B600, B1100, B2000

#### General Information

Fujitsu Microelectronics, Inc. offers a full line of low power, high speed bipolar gate arrays. All are fabricated using low power Schottky TTL technology with TTL-compatible inputs and outputs, and use NAND-type logic. The Fujitsu bipolar gate array family has two layer metallization for gate interconnect, automatic interconnect routing for guaranteed 90 percent cell utilization and proven device reliability. They all feature over 54 predesigned, hardwired logic cells, and over 50 predesigned 74 series MSI (softwired) macros. All Fujitsu bipolar gate arrays can accept user defined macros as well. To assure the success of your design, Fujitsu's gate arrays feature DC parametric and DC functional testing as well as AC parametric testing.

#### The Fujitsu TTL Gate Array Family

Device Name	B240	B350	B600	B1100
Device Numbering	MB111K Series	MB112K Series	MB113K Series	MB114K Series
Technology	LSTTL	LSTTL	LSTTL	LSTTL
Internal Cells	240 3-Input NAND	360 3-input NAND	616 3-Input NAND	1120 3-Input NAND
Cell Speed	1.9ns @ 0.8mW	1.9ns @ 0.8mW	1.9ns @ 0.8mW	1.9ns @ 0.8mW
Peripheral Buffers	40 40 Input Total 32 Output	48 28 Input Total 38 Output	64 64 Input Total 51 Output	88 88 Input Total 70 Output
Packaging	14, 16, 18, 20, 22, 24, 28, 40, 42-pin DIP	22, 24, 28, 40, 42, 48, 64-pin DIP 64-pin RIT (square PGA)	24, 28, 40, 42 48, 64-pin DIP 64, 88-pin RIT (square PGA)	64, 88, 107-pin RIT (square PGA)
Interface	Logic Diagram or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagram or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagram or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagram or FLDL (Net List) and FTDL (Formatted Test Data)
Macro Library	Same as B2000 description	Same as B2000 description	Same as B2000 description	Same as B2000 description
Turnaround Time	10 weeks after validation	10 weeks after validation	10 weeks after validation	10-weeks after validation
Price	Call Your FMI Sales Office	Call Your FMI Sales Office	Call Your FMI Sales Office	Call Your FMI Sales Office

B240  
B350  
B600  
B1100  
B2000

The Fujitsu TTL Gate Array  
Family, Continued

Device Name	B2000
Device Numbering	MB17K Series
Technology	Low Power Schottky TTL
Internal Cells	2108-3 input NAND
Cell Speed	1.5 ns typical
Peripheral Buffers	112 Input, Output or I/O 72 connected outputs maximum
Packaging	135-pin RIT (square)
Interface	FLDL (Net List) and FTDL (formatted test data)
Macro Library	54 predesigned hardwired logic cells; over 50 predesigned 74 Series MSI (softwired) macros; user defined macros allowed
Turnaround Time	16 weeks after validation
Price	Call your FMI Sales Office

DC and AC Characteristics

	B240	B350
DC Specifications	$V_{IH} = 2.0V$ Min. $V_{IL} = 0.8V$ Max. $I_{IH} = 20\mu A @ 2.4V$ $I_{IL} = -200\mu A @ 0.5V$ $V_{OH} = 2.4V$ Min. $V_{OL} = 0.5V$ Max. @ $I_{OL} 10mA$ $I_{OS} = -15$ to $-100mA$ $I_{OZ} = \pm 100\mu A$	$V_{IH} = 2.0V$ Min. $V_{IL} = 0.8V$ Max. $I_{IH} = 20\mu A @ 2.4V$ $I_{IL} = -200\mu A @ 0.5V$ $V_{OH} = 2.4V$ Min. $V_{OL} = 0.5V$ Max. @ $I_{OL} 10mA$ $I_{OS} = -15$ to $-100mA$ $I_{OZ} = \pm 100\mu A$
AC Specifications	Input Buffer 2.9ns, 1.15mW Output Buffer 4.5ns ( $C_L = 15pF$ ) 6.5ns ( $C_L = 50pF$ ) 4.9 mW	Input Buffer 2.9ns, 1.15mW Output Buffer 4.5ns ( $C_L = 15pF$ ) 6.5ns ( $C_L = 50pF$ ) 4.9 mW

	B600	B1100	B2000
DC Specifications	$V_{IH} = 2.0V$ Min. $V_{IL} = 0.8V$ Max. $I_{IH} = 20\mu A @ 2.4V$ $I_{IL} = -200\mu A @ 0.5V$ $V_{OL} = 0.5V$ Max. @ $I_{OL} 10mA$ $V_{OH} = 2.4$ Min. $I_{OS} = -15$ to $-100mA$ $I_{OZ} = \pm 100\mu A$	$V_{IH} = 2.0V$ Min. $V_{IL} = 0.8V$ Max. $I_{IH} = 20\mu A @ 2.4V$ $I_{IL} = -200\mu A @ 0.5V$ $V_{OH} = 2.4V$ Min. $V_{OL} = 0.5V$ Max. @ $I_{OL} 10mA$ $I_{OS} = -15$ to $-100mA$ $I_{OZ} = \pm 100\mu A$	Input $V_{IH} = 2.0V$ Min. Input $V_{IL} = 0.8V$ Max. Input $I_{IH} = 20\mu A$ . Max. Input $I_{IL} = -200\mu A$ . Max.  Output $V_{OL} = 0.5V @ 8mA$ Output $V_{OH} = 2.4V @ -2.6mA$ Output $I_{OZ} = \pm 100\mu A$ Output $I_{OS} = -10$ to $-90mA$
AC Specifications	Input Buffer 2.9ns, 1.15mW Output Buffer 4.5ns ( $C_L = 15pF$ ) 6.5ns ( $C_L = 50pF$ ) 4.9 mW	Input Buffer 2.9ns, 1.15mW Output Buffer 4.5ns ( $C_L = 15pF$ ) 6.5ns ( $C_L = 50pF$ ) 4.9 mW	Input Buffer 3.5ns typ. (NAND) Output Buffer 3.5ns typ. (NAND)

# CMOS Gate Arrays

# FUJITSU

## ■ C440, C770, C1275 C2000, C2600, C3900, C8000

### General Information

Fujitsu offers a full line of very highly integrated, low power high speed gate arrays which are fabricated with silicon gate CMOS technology. Fujitsu also guarantees 90 percent cell utilization for fully automatic placement and routing.

The Fujitsu CMOS Gate Array family offers more than 100 macro elements and contains gate arrays ranging in size from 400 basic cells to 8000 basic cells (2-input NAND equivalents).

All use Fujitsu's sophisticated computer-aided design system, which fully automates both the design layout and fabrication. The designs are implemented on-chip by interconnecting basic cells and I/O cells using a double layer metallization process.

To assure the success of your design, Fujitsu's gate arrays feature DC parametric and DC functional testing as well as AC parametric testing.

### The Fujitsu CMOS Gate Array Family, H Series

Device Name	C440H	C770H	C1275H	C2000H	C3900H
Device Numbering	MB64HK Series	MB62HK Series	MB63HK Series	MB60HK Series	MB61HK Series
Internal Cells	440-2 Input	770-2 Input	1275-2 Input	2000-2 Input	3900-2 Input
Cell Speed	4.0nS	4.0nS	4.0nS	4.0nS	4.0nS
Peripheral Buffer	54 Inputs or Outputs	74 Inputs or Outputs	80 Inputs or Outputs	68 Inputs or Outputs	68 Inputs or Outputs
Packaging	16, 18, 20, 22, 24, 28, 40, 42-pin DIP 64-pin RIT (square PGA)	24, 28, 40, 42, 48, 64-pin DIP 64-pin RIT (square PGA)	28, 40, 42, 48, 64-pin DIP 64-pin RIT (square PGA)	40, 42, 48-pin DIP 64, 88-pin RIT (square PGA)	64, 68-pin RIT (square PGA)
Interface	Logic Diagrams or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagrams or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagrams or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagrams or FLDL (Net List) and FTDL (Formatted Test Data)	Logic Diagrams or FLDL (Net List) and FTDL (Formatted Test Data)
Turnaround Time	14-weeks after validation	14-weeks after validation	14-weeks after validation	14-weeks after validation	14-weeks after validation
Price	Call Your FMI Sales Office	Call Your FMI Sales Office	Call Your FMI Sales Office	Call Your FMI Sales Office	Call Your FMI Sales Office

C440, C770, C1275, C2000  
C2600, C3900, C8000

The Fujitsu CMOS Gate  
Array Family, VH Series

Device Name	C2600VH	C3900VH	C8000VH
Device Numbering	MB60VHK Series	MB61VHK Series	MB66VHK Series
Internal Cells	2600 2-Input NAND	3900 2-Input NAND	8000 2-Input NAND
Cell Speed	2.5ns	2.5ns	2.5ns
Peripheral Buffers	106 Input/Output	127 Input/Output	160 Input/Output
Packaging	28, 40, 42, 48, 64-pin DIP 48, 64, 80-pin Flatpack 64, 88, 135-pin RIT (square PGA) 48, 64-pin LCC	40, 42, 48, 64-pin DIP 64, 88, 135-pin RIT (square PGA) 48, 64-pin LCC 48, 64, 80-pin Flatpack	64, 135, 179-pin RIT (square PGA)

DC Electrical Characteristics  
(All CMOS Arrays)

Parameter	Min	Typ	Max	Unit
Supply Voltage ( $V_{DD}$ )	4.25	5.0	5.25	V
Input High Voltage ( $V_{IH}$ )	2.0	—	—	V
Input Low Voltage ( $V_{IL}$ )	$V_{SS}$	—	0.8	V
Input High Voltage at $I_{OH} = -0.4mA$	4.2	—	$V_{DD}$	V
Output Low Voltage at $I_{OH} = 20mA$	$V_{SS}$	—	0.4	V

# Gate Arrays

# FUJITSU

## ■ Packaging, About Fujitsu Microelectronics, Inc.

### Package Information

Fujitsu has a wide variety of standard device packaging available for your gate array design as shown below. If you have packaging requirements that are not on this table (Flat Pack, LCC), a Fujitsu applications engineer will be happy to discuss your needs.

### Packaging Options

	B240	B350	B600	B1100	B2000	C440H	C770H	C1275H	C2000H	C3900H	C2600VH	C3900VH	C8000VH
DIP-14	■												
DIP-16	● ■					● ■							
DIP-18	● ■					● ■							
DIP-20	● ■					● ■							
DIP-22	● ■	● ■				● ■							
DIP-24	● ■	● ■	● ■			● ■	● ■						
DIP-28	● ■	● ■	● ■			● ■	● ■	● ■			■		
DIP-40	● ■	● ■	● ■			● ■	● ■	● ■	●		● ■	●	
DIP-42	● ■	● ■	● ■			■	● ■	● ■	●		● ■	●	
DIP-48		●	●				●	●	●		●	●	
DIP-64							■	■					
SDIP-64		■	■								■	■	
FPT-48											■	■	
FPT-64											■	■	
FPT-80											■	■	
RIT-64		●	●	●		●	●	●	●	●	●	●	●
RIT-88			●	●					●	●	●	●	
RIT-107*			●										
RIT-135					●						●	●	●
RIT-179													●

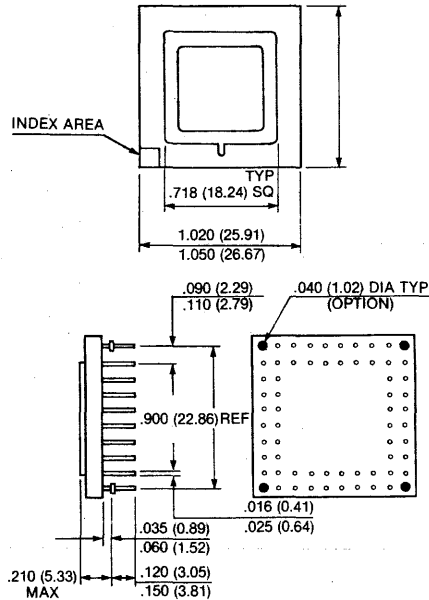
\* RIT-107 is under development.

- Ceramic
- Plastic

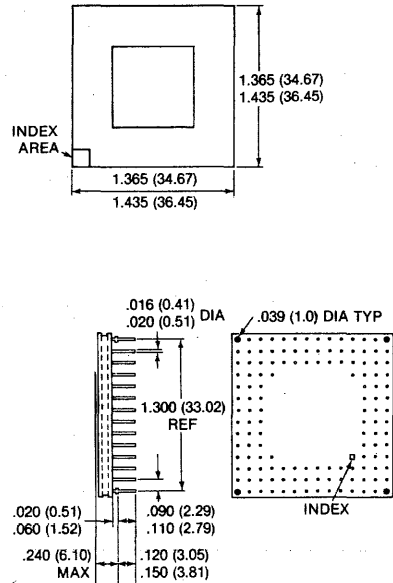


**(RIT) Pin Square  
Package Dimensions**  
Dimension in inches  
(millimeters)

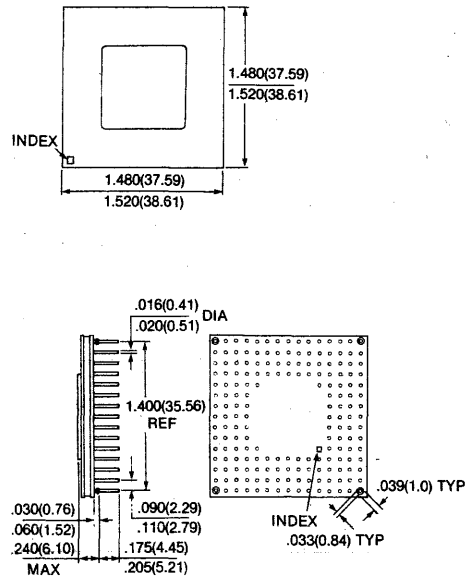
**64-Pin Square  
Ceramic Package**



**135-Pin Square  
Ceramic Package**



**179-Pin Square  
Ceramic Package**



**About Fujitsu  
Microelectronics, Inc.**

**Fujitsu Microelectronics, Inc.—A U.S. Organization**  
Fujitsu Microelectronics, Inc. (FMI) is a U.S. subsidiary of Fujitsu Limited of Tokyo. A California Corporation, FMI is responsible for the marketing and sales of all semiconductor products in North, Central and South America.

Fujitsu Limited manufactures and markets advanced data processing and telecommunications systems, semiconductors and electronic components on a world wide scale. Fujitsu limited is ranked as Japan's number one computer manufacturer with sales in the \$2 billion range.

**A Leader in Large Scale Logic**

Since 1974, Fujitsu Microelectronics has executed over 3,000 gate array designs. Gate arrays from Fujitsu give your designs a significant performance edge. In 1982, we introduced our CMOS device with 8,000 gates and 2.5 ns delay times. Our leadership Schottky TTL gate arrays is the first with 2,000 gates and 0.95 ns delays. Our leadership ECL arrays, with 400-picosecond delays are the fastest production arrays in the world.

Fujitsu's logic products offer leading edge technology, a complete selection of both bipolar and si-gate CMOS

devices; and experience that can serve you through the coming generations of logic products.

**About Fujitsu  
Microelectronics, Inc.**

Fujitsu Microelectronics has completed a new assembly and test facility in order to better service our North American customers. The 66,000-square-foot facility is located in Kearny Mesa Industrial Park near downtown San Diego. The building was dedicated in June 1981 and is now fully operational.

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## ARINC 429 Bus Interface Line Driver Circuit

### Features

- INPUTS T<sup>2</sup>L AND CMOS COMPATIBLE
- ADJUSTABLE RISE AND FALL TIMES VIA 2 EXTERNAL CAPACITORS
- PROGRAMMABLE OUTPUT DIFFERENTIAL RANGE VIA VOLTAGE REFERENCE INPUT (V<sub>REF</sub>)
- POWER STROBE INPUT PERMITS LOW QUIESCENT POWER OF < 20mW
- OUTPUTS ARE INHIBITED (0 VOLTS) IF DATA (A) AND DATA (B) INPUTS ARE BOTH IN THE "LOGIC ONE" STATE
- CAN OPERATE UP TO A 100 KILOBITS DATA RATE
- OUTPUT SHORT CIRCUIT PROOF AND CONTAINS OVERVOLTAGE PROTECTION
- DATA "A" AND DATA "B" SIGNALS ARE "AND'D" WITH CLOCK AND SYNC SIGNALS
- FULL MILITARY TEMPERATURE RANGE

### Description

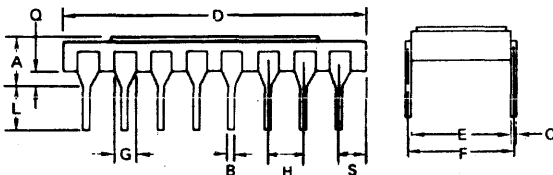
The HS-3182 ARINC 429 bus interface driver circuit is a monolithic dielectrically isolated bipolar differential line driver designed to meet the specifications of ARINC 429. This device is intended to be used with a companion chip, HS-3282 CMOS ARINC bus interface circuit, which provides the data formatting and processor interface function.

All logic inputs are T<sup>2</sup>L and CMOS compatible. In addition to the DATA(A) and DATA(B) inputs there are also inputs for a CLOCK and SYNC signal which are AND'D with the DATA inputs. This feature was added to enhance system performance and to allow the HS-3182 to be used with devices other than the HS-3282. Also adding to system performance is the STROBE input. To minimize power consumption the STROBE input can be asserted to place the chip in the power-down mode where it draws substantially less current. Four power supplies are required; +V = +15V ± 10%, -V = -15V ± 10%, V<sub>1</sub> = 5V ± 5% and V<sub>REF</sub>. V<sub>REF</sub> is used to program the output voltage swing, such that V<sub>OUT</sub> (DIFF) = ± 2V<sub>REF</sub>. Typically, V<sub>REF</sub> = V<sub>1</sub> = 5V ± 5%.

The driver output impedance is 75Ω ± 20% at 25°C. Output rise and fall times are programmed through the use of two external capacitors, C<sub>A</sub> and C<sub>B</sub>. To meet the requirements for rise and fall times as specified in ARINC 429, C<sub>A</sub> = C<sub>B</sub> = 75pF for the high speed operation (100 KBPS) and 500 pF for the low speed operation (12-14.5 KBPS). The outputs are protected against overvoltage and short circuit as shown in the Block Diagram. This device is designed to operate with a case temperature range of -55°C to +125°C.

### Package

16 LEAD BRAZED DIP



LEAD COUNT	DIM. A	DIM. B	DIM. C	DIM. D	DIM. E	DIM. F	DIM. G	DIM. H	DIM. L	DIM. Q	DIM. S
16	.200	.014 .023	.008 .015	.840	.220 .310	.290 .320	.030 .070	.100 .085C	.125 .200	.015 .060	.060

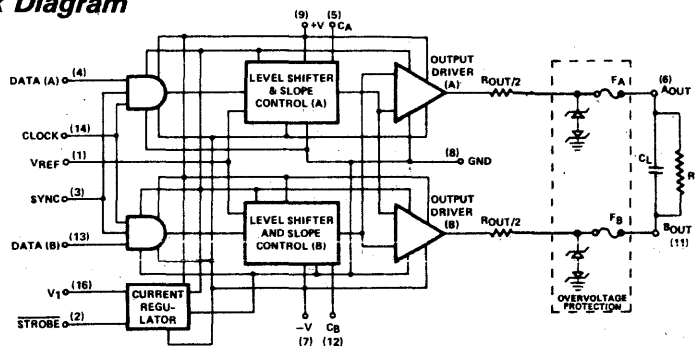
MIN.  
MAX.

NOTE: DIMENSIONS IN INCHES

### Truth Table

STROBE	SYNC	CLOCK	DATA (A)	DATA (B)	A <sub>OUT</sub>	B <sub>OUT</sub>	COMMENTS
H	X	X	X	X	HI-Z	HI-Z	Power-Down State
L	X	L	X	X	OV	OV	NULL
L	L	X	X	X	OV	OV	NULL
L	H	H	L	L	OV	OV	NULL
L	H	H	L	H	-V <sub>REF</sub>	+V <sub>REF</sub>	LOW
L	H	H	H	L	+V <sub>REF</sub>	-V <sub>REF</sub>	HIGH
L	H	H	H	H	OV	OV	NULL

### Block Diagram



### Pinout TOP VIEW

V <sub>REF</sub>	1	16	V <sub>1</sub>
STROBE	2	15	NC
SYNC	3	14	CLOCK
DATA (A)	4	13	DATA (B)
C <sub>A</sub>	5	12	C <sub>B</sub>
A <sub>OUT</sub>	6	11	B <sub>OUT</sub>
-V	7	10	NC
GND	8	9	+V



**ABSOLUTE MAXIMUM RATINGS**

Voltage Between +V and -V terminals	40V
V <sub>1</sub>	7V
V <sub>REF</sub>	6V
Logic Input Voltage	(Gnd - 0.3V) to (V <sub>1</sub> + 0.3V)
Output Short Circuit Duration	See Note 1
Output Overvoltage Protection	See Note 3

**OPERATING RANGE**

Operating Voltage: +V	+15V ± 10%
-V	-15V ± 10%
V <sub>1</sub>	5V ± 5%
V <sub>REF</sub> (for ARINC 429)	5V ± 5%
V <sub>REF</sub> (applications other than ARINC)	0V to 6V
Operating Case Temperature	-55 °C to +125 °C
Storage Temperature	-65 °C to +150 °C

**ELECTRICAL CHARACTERISTICS**

+V = +15V ± 10%, -V = -15V ± 10%, V<sub>1</sub> = V<sub>REF</sub> = 5V ± 5%  
Case Temperature: -55 °C to +125 °C

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNITS	TEST CONDITION
I <sub>CCPD</sub> (+V)	Supply Current +V (Power Down)	-475	+10	475	μA	STROBE = H
I <sub>CCPD</sub> (-V)	Supply Current -V (Power Down)	-475	+10	475	μA	STROBE = H
I <sub>CCOP</sub> (+V)	Supply Current +V (Operating)		11	16	mA	No Load (0-100 kBps)
I <sub>CCOP</sub> (-V)	Supply Current -V (Operating)		-10	-16	mA	No Load (0-100 kBps)
I <sub>CCOP</sub> (V <sub>1</sub> )	Supply Current V <sub>1</sub> (Operating)		600	975	μA	No Load (0-100 kBps)
I <sub>CCOP</sub> (V <sub>REF</sub> )	Supply Current V <sub>REF</sub> (Operating)	-1.0	-0.4	-0.15	mA	No Load (0-100 kBps)
V <sub>IH</sub>	Logic "1" Input Voltage	2.0			V	
V <sub>IL</sub>	Logic "0" Input Voltage			0.6	V	
V <sub>OH</sub>	Output Voltage High (Output to ground)	V <sub>REF</sub> (-250 mV)	V <sub>REF</sub>	V <sub>REF</sub> (+250 mV)		No Load (0-100 kBps)
V <sub>OL</sub>	Output Voltage Low (Output to ground)	-V <sub>REF</sub> (-250 mV)	-V <sub>REF</sub>	-V <sub>REF</sub> (+250 mV)		No Load (0-100 kBps)
V <sub>NULL</sub>	Output Voltage Null	-250	0	+250	mV	No Load (0-100 kBps)
I <sub>IL</sub>	Input Current (Input Low)		-1	-20	μA	
I <sub>IH</sub>	Input Current (Input High)		1	10	μA	
I <sub>OHSC</sub>	Output Short Circuit Current (Output High)	-80	-100		mA	Short to GND
I <sub>OLSC</sub>	Output Short Circuit Current (Output Low)	80	100		mA	Short to GND
C <sub>IN</sub>	Input Capacitance			15	pF	
I <sub>SC</sub> (+V)	Supply Current +V (Short Circuit)			150	mA	Short to GND
I <sub>SC</sub> (-V)	Supply Current -V (Short Circuit)			-150	mA	Short to GND

NOTES:

- ① Heat sink may be required for 100KBPS @ +125°C and output short circuit @ +125°C.  
Thermal Characteristics: T<sub>CASE</sub> = T<sub>(Junction)</sub> - θ<sub>(Junction - Case)</sub> P<sub>(Dissipation)</sub>.

Where: T<sub>(Junction Max)</sub> = +175°C  
θ<sub>(Junction - Case)</sub> = 19.6°C/W  
θ<sub>(Junction - Ambient)</sub> = 86.5°C/W

- ② Full Load for ARINC 429: R<sub>L</sub> = 400Ω and C<sub>L</sub> = 30,000pF in parallel between A<sub>OUT</sub> and B<sub>OUT</sub>. (See Block Diagram).

- ③ Output Overvoltage Protection: The fuses used for output overvoltage protection may be blown by a fault at each output of greater than ±6.5V relative to GND.

**POWER SPECIFICATIONS**

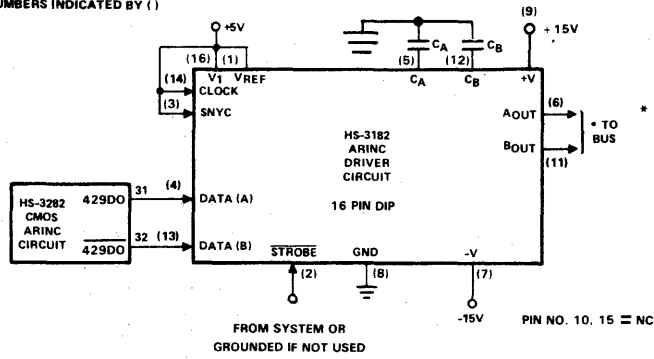
**NOMINAL POWER @25 °C, +V = 15V, -V = -15V, V<sub>1</sub> = V<sub>REF</sub> = 5V**

DATA RATE (KBPS)	LOAD	+V	-V	V <sub>1</sub>	CHIP POWER	POWER DISSIPATION IN LOAD
0-100	No Load	11 mA	-10 mA	600 μA	325 mW	0
12.5-14	Full Load*	24 mA	-24 mA	600 μA	660 mW	60 mW
100	Full Load*	46 mA	-46 mA	600 μA	1 watt	325 mW

\* See Note (2) above.

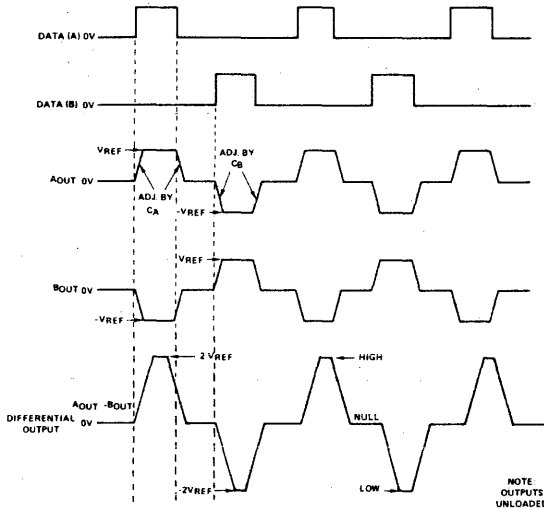
### Typical Application

PIN NUMBERS INDICATED BY ( )



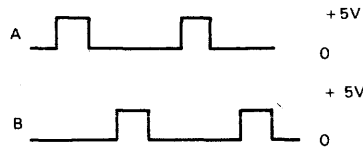
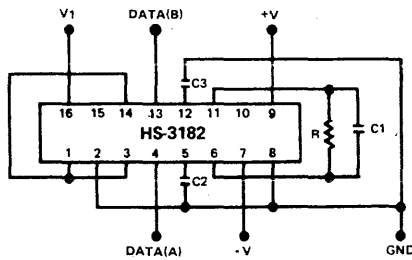
\* The rise and fall time of the outputs are set to ARINC specified values by  $C_A$  and  $C_B$ . Typical  $C_A = C_B = 75\text{pF}$  for high speed and  $500\text{pF}$  for low speed operation. The output HI and low levels are set to ARINC specifications by  $V_{REF}$ .

### Driver Waveforms



When the Data (A) input is in the Logic One state and the Data (B) input is in the Logic Zero state, A Out is equal to  $V_{REF}$  and B Out is equal to  $-V_{REF}$  this constitutes the Output High state. Data (A) and Data (B) both in the Logic Zero state causes both A Out and B Out to be equal to  $0V$  which designates the output Null state. Data (A) in the Logic Zero state and Data (B) in the Logic One state cause A Out to be equal to  $-V_{REF}$  and B Out to be equal to  $V_{REF}$  which is the Output Low state. Both A Out and B Outputs are high impedance when the transmitter is disabled via the power strobe input.

### Burn-In Schematic



- R =  $400\Omega$ , 1/4 watt.
- C1 =  $0.03\mu\text{F}$
- C2 = C3 =  $500\text{pF}$ , NPO.
- +V =  $\pm 15V \pm 10\%$ .
- V =  $\pm 15V \pm 10\%$ .
- V<sub>1</sub> =  $\pm 5V \pm 10\%$ .

- Ambient Temp. Max =  $+125^\circ\text{C}$
- Package = 16 Lead Brazed Dip
- Pulse Conditions =
- A & B =  $6.25\text{KHz}$ , 25% on & 75% off duty cycle.
- B is delayed one-half cycle and in sync with A.

A  $0.01\text{F}$  decoupling capacitor is required on each of the three supply lines (+V, -V, and V<sub>1</sub>) at every 3rd Burn-in socket.

**HARRIS SEMICONDUCTOR PRODUCT FLOW  
MIL-STD-883, METHOD 5004 CLASS B**

**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual.	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: <b>A</b> Fine <b>B</b> Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

**Note:**

- Traceability:** All devices are assigned date code identification that provides traceability back to the inspection lot.
- Branding:** All devices are branded with the part number and EIA date code.
- Aged Products:** Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
- Additional Requirements:** Sample Group A electrical tests are performed on a lot acceptance basis.

**UNITED STATES**

**EASTERN REGION**

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
TWX: 710-332-1074

106 Seventh Street  
Garden City, NY 11530  
(516) 747-6776

**WESTERN REGION**

1503 South Coast Drive  
Suite 320  
Costa Mesa, CA 92626  
(714) 957-6557  
TWX: 910-595-1533

625 Ellis, Suite 300  
Mt. View, CA 94043  
(415) 964-6443  
TWX: 910-379-6431

**MIDWEST REGION**

2850 Metro Dr., Suite 703  
Minneapolis, MN 55420  
(612) 854-3224  
TWX: 910-576-3418

**HOME OFFICE**

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

**EUROPEAN**

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



**HARRIS**

**CUSTOM INTEGRATED CIRCUITS DIVISION**

Harris Semiconductor

CUSTOM/SEMICUSTOM

### Description

The Harris HS-3282 is a high performance CMOS bus interface circuit that is intended to meet the requirements of ARINC Specification 429, and similar encoded, time multiplexed serial data protocols. The ARINC 429 bus interface circuit consists of two (2) receivers and a transmitter operating independently as shown in Figure 1. The two receivers operate at a frequency that is ten (10) times the receiver data rate, which can be the same or different from the transmitter data rate. Although the two receivers operate at the same frequency, they are functionally independent and each receives serial data asynchronously. The transmitter section of the ARINC bus interface circuit consists mainly of a First-In First-Out (FIFO) memory and timing circuit. The FIFO memory is used to hold eight (8) ARINC data words for transmission serially. The timing circuit is used to correctly separate each ARINC word as required by ARINC Specification 429.

Even though ARINC Specification 429 specifies a 32-bit word, including parity, the HS-3282 can be programmed to also operate with a word length of 25 bits. The incoming receiver data word parity is checked, and a parity status is stored in the receiver latch and is outputted on Pin BD08 during the 1st word. [A logic "0" indicates that an odd number of logic "1"s were received and stored; a logic "1" indicates that an even number of logic "1"s were received and stored]. In the transmitter the parity generator will generate either odd or even parity depending upon the status of PARCK control signal. A logic "0" on BD12 will cause odd parity to be generated and inputted to the output data stream. Conversely, a logic "1" on BD12 will result in the generation of even parity that will be inputted to the output data stream.

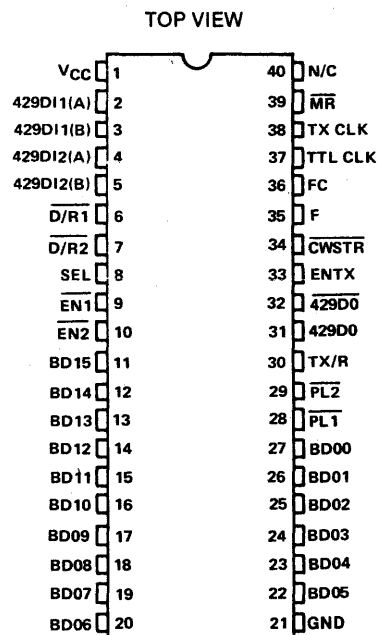
More versatility is provided in both the transmitter and receiver by the addition of an external TTL clock input allowing the bus interface circuit to operate at data rates from 0 to 0.1 megabits. The TTL external clock must be ten (10) times the data rate to insure no data ambiguity.

The ARINC bus interface circuit is fully guaranteed to support the data rates of ARINC specification 429 over both the voltage ( $\pm 5\%$ ) and full military temperature range. It interfaces with TTL, CMOS or NMOS support circuitry, and uses the standard 5-volt VCC supply.

### Features

- ARINC SPECIFICATION 429 COMPATIBLE
- DATA RATES OF 100 KILOBITS OR 12.5 KILOBITS
- SEPARATE RECEIVER AND TRANSMITTER SECTION
- DUAL AND INDEPENDENT RECEIVERS, CONNECTING DIRECTLY TO ARINC BUS
- SERIAL TO PARALLEL RECEIVER DATA CONVERSION
- PARALLEL TO SERIAL TRANSMITTER DATA CONVERSION
- WORD LENGTHS OF 25 OR 32 BITS
- PARITY STATUS OF RECEIVED DATA
- GENERATE PARITY OF TRANSMITTER DATA
- AUTOMATIC WORD GAP TIMER
- SINGLE 5-VOLT SUPPLY
- LOW POWER DISSIPATION
- FULL MILITARY TEMPERATURE RANGE

### Pinout



# Specifications HS-3282



## ABSOLUTE MAXIMUM RATINGS

Voltage at any Pin (Except 2, 3, 4, & 5)	-0.3V to V <sub>CC</sub> + 0.3V
Voltage at Pins 2, 3, 4, & 5	-29V to +29V
Maximum V <sub>CC</sub>	7.0V
Operating Temperature Range	-55°C to +125°C
Storage Temperature	-65°C to +150°C

**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

## ELECTRICAL CHARACTERISTICS

V<sub>CC</sub> = 5V ± 5% TA = Operating Temperature Range

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	TEST CONDITIONS
<u>ARINC INPUTS</u>						
V <sub>IH</sub>	Logic "1" Input Voltage	6.7	10	13	V	Pin 2-3, 4-5
V <sub>IL</sub>	Logic "0" Input Voltage	-6.7	-10	-13	V	Pin 2-3, 4-5
V <sub>IN</sub>	Null Input Voltage	-2.5	0	+2.5	V	
V <sub>CH</sub>	Common Mode V			±5	V	
I <sub>IH</sub>	Input Leakage	75	135	200	μA	V <sub>IN</sub> = V <sub>CC</sub>
I <sub>IL</sub>	Input Leakage	-200	-325	-450	μA	V <sub>IN</sub> = 0
R <sub>I</sub>	Differential Input Impedance	6			KΩ	
R <sub>H</sub>	Input Impedance to V <sub>CC</sub>	12			KΩ	
R <sub>G</sub>	Input Impedance to GND	12			KΩ	
C <sub>I</sub>	Differential Input Capacitance*			20	pF	
C <sub>H</sub>	Input Capacitance to V <sub>CC</sub> *			20	pF	
C <sub>G</sub>	Input Capacitance to GND*			20	pF	
<u>BI-DIR. INPUTS</u>						
V <sub>IH</sub>	Logic "1" Input Voltage	2.1			V	
V <sub>IL</sub>	Logic "0" Input Voltage			0.7	V	
I <sub>I</sub>	Input Leakage	-1.5		-1.5	μA	0 < V <sub>IN</sub> < V <sub>CC</sub>
<u>ALL OTHER INPUTS</u>						
V <sub>IH</sub>	Logic "1" Input Voltage	3.5			V	
V <sub>IL</sub>	Logic "0" Input Voltage			0.7	V	
I <sub>IH</sub>	Input Leakage (Except Pin 36)			10	μA	V <sub>IN</sub> = V <sub>CC</sub>
	Input Leakage (Pin 36)			50	μA	V <sub>IN</sub> = V <sub>CC</sub>
I <sub>IL</sub>	Input Leakage			-10	μA	V <sub>IN</sub> = 0
C <sub>I</sub>	Input Capacitance*			15	pF	
<u>OUTPUTS</u>						
(Including Bi-directional Outputs)						
V <sub>OH</sub>	Logic "1" Output	2.7			V	I <sub>OH</sub> = -1.5mA
V <sub>OL</sub>	Logic "0" Output			0.4	V	I <sub>OL</sub> = 1.8mA
C <sub>O</sub>	Output Capacitance*			15	pF	
I <sub>CC1</sub>	Supply Current Stand-by			20.0	mA	V <sub>CC</sub> = 5.5, V <sub>IN</sub> = 0
I <sub>CC2</sub>	Supply Current Operation			20.0	mA	

\*Guaranteed but not tested.



**ELECTRICAL CHARACTERISTICS (Continued)**

SYMBOL	PARAMETER	DATA RATE = 100 KBFS		DATA RATE = 12.5 KBFS		UNITS	TEST CONDITIONS	
		MIN	MAX	MIN	MAX			
FC <sup>1</sup>	Clock Frequency		1		1	MHz	≈50% Duty Cycle	
FD	Data Rate		100		12.5	KHz		
TLHC	TTL Clock Rise Time		10		10	ns		
THLC	TTL Clock Fall Time		10		10	ns		
TMR	Master Reset Pulse Width	200		200		ns		
<b>RECEIVER TIMING</b>								
TD/R <sup>2</sup>	Receiver Device Ready Time From 32nd Data Bit		16		128	μs		
TD/REN	Device Ready to Enable Time	0		0		ns		
TEN	Data Enable Pulse Width	200		200		ns		
TENEN	Data Enable to Data Enable Time	50		50		ns		
TEND/R	Data Enable to Device Ready Reset Time		200		200	ns		
TENDATA	Output Data Valid to Enable Time		200		200	ns		
TENSEL	Data Enable to Data Select Time	20		20		ns		
TSELEN	Data Select to Data Enable Time	20		20		ns		
TDATAEN	Output Data Disable Time		30		30	ns		
<b>CONTROL WORD TIMING</b>								
TCWSTR	Control Word Register Strobe Pulse Width	130		130		ns		
TCWSET	Control Word Setup Time	130		130		ns		
TCWHLD	Control Word Hold Time	0		0		ns		
<b>TRANSMITTER FIFO WRITE TIMING</b>								
TPL	Parallel Load Pulse Width	200		200		ns		
TPL12	Parallel Load 1 to Parallel Load 2 Delay	0		0		ns		
TTX/R	Transmitter Ready Delay Time		840		840	ns		
TDWSET	Data Word Setup Time	110		110		ns		
TDWHLD	Data Word Hold Time	0		0		ns		
<b>TRANSMITTER OUTPUT TIMING</b>								
TENDAT	Enable Transmit to Output Data Valid Time		25		200	μs		
TBIT	Output Data BIT Time	4.95	5.05	39.6	40.4	μs		
TNUL	Output Data Nul Time	4.95	5.05	39.6	40.4	μs		
TDTX/R	Data Transmission Word to TX/R Set Time		50		50	ns		
TENTX/R	Enable Transmit Turn Off Time	0		0		ns		
TGAP	Data Word Gap Time	39.6	40.4	316.8	323.2	μs		
<b>REPEATER OPERATION TIMING</b>								
TENPL	Data Enable to Parallel Load Delay Time	0		0		ns		
TPLEN	Data Enable Hold For Parallel Load Time	0		0		ns		
TTX/REN	Enable Transmit Delay Time	0		0		ns		

1. 60-40 Duty Cycle Acceptable      2. Same Delay For 25 bit Word Format

CUSTOMER SERVICE DEPARTMENT

CUSTOMER SERVICE DEPARTMENT

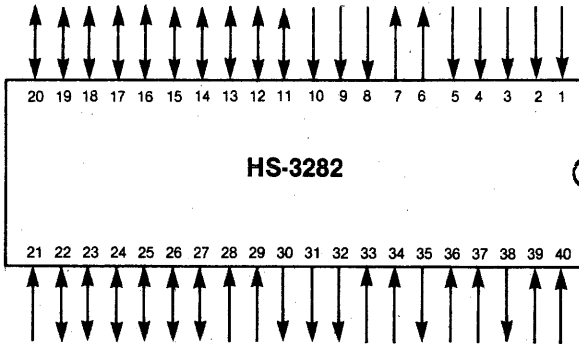
# Pin Assignments



PIN	SYMBOL	SECTION	DESCRIPTION
1	VCC	Receiver/ Transmitter	Supply pin. 5 volts $\pm$ 5%.
2	429 DI1 (A)	Receiver	ARINC 429 data input to Receiver 1.
3	429 DI1 (B)	Receiver	ARINC 429 data input to Receiver 1.
4	429 DI2 (A)	Receiver	ARINC 429 data input to Receiver 2.
5	429 DI2 (B)	Receiver	ARINC 429 data input to Receiver 2.
6	$\overline{D/R1}$	Receiver	Device ready flag output from Receiver 1 indicating a valid data word is ready to be fetched.
7	$\overline{D/R2}$	Receiver	Device ready flag output from Receiver 2 indicating a valid data word is ready to be fetched.
8	SEL	Receiver	Bus Data Selector - input signal to select one of two 16-bit words from either Receiver 1 or 2.
9	$\overline{EN1}$	Receiver	Input signal to enable data from Receiver 1 onto the data bus.
10	$\overline{EN2}$	Receiver	Input signal to enable data from Receiver 2 onto the data bus.
11	BD15	Receiver/ Transmitter	Bi-directional data bus for fetching data from either of the Receivers, or for loading data into the Transmitter memory or control word register. See Control Word Table for description of Control Word bits.
12	BD14	Receiver/ Transmitter	See pin 11.

PIN	SYMBOL	SECTION	DESCRIPTION
13	BD13	Receiver/ Transmitter	See pin 11.
14	BD12	Receiver/ Transmitter	See pin 11.
15	BD11	Receiver/ Transmitter	See pin 11.
16	BD10	Receiver/ Transmitter	See pin 11.
17	BD09	Receiver/ Transmitter	See pin 11.
18	BD08	Receiver/ Transmitter	See pin 11.
19	BD07	Receiver/ Transmitter	See pin 11.
20	BD06	Receiver/ Transmitter	See pin 11.
21	GND	Receiver/ Transmitter	Circuit ground.
22	BD05	Receiver/ Transmitter	See pin 11.
23	BD04	Receiver/ Transmitter	See pin 11. Control Word function not applicable.
24	BD03	Receiver/ Transmitter	See pin 11. Control Word function not applicable.
25	BD02	Receiver/ Transmitter	See pin 11. Control Word function not applicable.
26	BD01	Receiver/ Transmitter	See pin 11. Control Word function not applicable.
27	BD00	Receiver/ Transmitter	See pin 11. Control Word function not applicable.
28	$\overline{PL1}$	Transmitter	Parallel load input signal loading the first 16-Bit word into the Transmitter memory.

## Pinout



PIN	SYMBOL	SECTION	DESCRIPTION
29	$\overline{PL2}$	Transmitter	Parallel load input signal loading the second 16-Bit word into the Transmitter memory and initiates data transfer into the memory stack.
30	TX/R	Transmitter	Transmitter flag output to indicate the memory is empty.
31	429D0	Transmitter	Data output from Transmitter.
32	429D0	Transmitter	Data output from Transmitter.
33	ENTX	Transmitter	Transmitter Enable Input signal to initiate data transmission from FIFO memory.
34	$\overline{CWSTR}$	Receiver/ Transmitter	Control word input strobe signal to latch the control word from the data bus into the control word register.
35	NC	Receiver/ Transmitter	This pin should be left open.

PIN	SYMBOL	SECTION	DESCRIPTION
36	NC	Receiver/ Transmitter	No connect. This pin may be left open or tied low but never tied high.
37	TTLCLK	Receiver/ Transmitter	External clock input. May be either ten (10) or eighty (80) times the data rate. If using both ARINC data rates it must be ten (10) times the highest data rate. (Typically 1 MHz). If using crystal oscillator, this pin may be left open or tied high but never tied low.
38	TXCLK	Transmitter	Transmitter Clock output. Delivers a clock frequency equal to the transmitter data rate.
39	$\overline{MR}$	Receiver/ Transmitter	Master Reset. Active low pulse used to reset FIFO, bit counters, gap timer, word count signal, TX/R and various other flags and controls. Usually only used on Power-Up or System Reset.
40	—	—	No connection.



The HS-3282 is designed to support ARINC Specification 429 and other serial data protocols that use a similar format by collecting the receiving, transmitting, synchronizing, timing and parity functions on a single, low power LSI circuit. It goes beyond the ARINC requirements of providing for either odd or even parity, and giving the user a choice of either 25 or 32 bit word lengths. The receiver and transmitter sections operate independently of each other. The serial-to-parallel conversion required of the receiver and the parallel-to-serial conversion requirements of the transmitter have been incorporated into the bus interface circuit.

Provisions have been made through the addition of an external TTL clock input to provide data rate flexibility. This requires an external TTL clock that is 10 times the data rate.

To obtain the flexibility discussed above, a number of external control signals are required. To reduce the pin count requirements, an internal control word register is used. The control word is latched from the data bus into the register by the Control Word Strobe (CWSTR) signal going to a logic "0". Eleven (11) control functions are used, and along with the bus data (BD) line are listed below:

**CONTROL WORD**

PIN NAME	SYMBOL	FUNCTION
BD05	SLFTST	Connects the self test signal from the transmitter directly to the receiver shift registers, bypassing the input receivers. Receiver 1 receives Data true and Receiver 2 receives Data not. Note that the transmitter output remains active. (Logic "0" on SLFTST Enables Self Test).
BD06	SDENB1	Signal to activate the Source Destination (SD) Decoder for Receiver 1. (Logic "1" activates SD Decoder).
BD07	X1	If SDENB1 = "1" then this bit is compared with ARINC Data Bit #9. If X1 also matches (see Y1), the word will be accepted by the Receiver 1. If SDENB1 = "0" this bit becomes a don't care.
BD08	Y1	If SDENB1 = "1" then this bit is compared with ARINC Data Bit #10. If X1 also matches (see X1), the word will be accepted by the Receiver 1. If SDENB1 = "0" this bit becomes a don't care.
BD09	SDENB2	Signal to activate the Source Destination (SD) Decoder for Receiver 2. (Logic "1" activates SD Decoder).
BD10	X2	If SDENB2 = "1" then this bit is compared with ARINC Data Bit #9. If Y2 also matches (see Y2), the word will be accepted by the Receiver 2. If SDENB2 = "0" this bit becomes a don't care.
BD11	Y2	If SDENB2 = "1" then this bit is compared with ARINC Data Bit #10. If X2 also matches (see X2), the word will be accepted by the Receiver 2. If SDENB2 = "0" this bit becomes a don't care.
BD12	PARCK	Signal used to invert the transmitter parity bit for test of parity circuits. Logic "0" selects normal odd parity. Logic "1" selects even parity.
BD13	TXSEL	Selects high or low Transmitter data rate. If TXSEL = "0" then transmitter data rate is equal to the clock rate divided by ten (10). If TXSEL = "1" then transmitter data rate is equal to the clock rate divided by eighty (80).
BD14	RCVSEL	Selects high or low Receiver data rate. If RCVSEL = "0" then the received date rate should be equal to the clock rate divided by ten (10). If RCVSEL = "1" then the received data rate should be equal to the clock rate divided by eighty (80).
BD15	WLSEL	Selects word length. If WLSEL = "0" a 32-bit word format will be selected. If WLSEL = "1" a 25-bit word format will be selected.

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## Operational Description (Continued)



ARINC 429 DATA FORMAT as input to the Receiver and output from the Transmitter is as follows:

TABLE 1 - ARINC 429 DATA FORMAT

ARINC BIT #	FUNCTIONS
1 - 8	Label
9 - 10	SDI or Data
11	LSB
12 - 27	Data
28	MSB
29	Sign
30, 31	SSM
32	Parity

This format is shuffled when seen on the sixteen bi-directional input/outputs. The format shown below is used from the receivers and input to the transmitter:

TABLE 2 - WORD 1 FORMAT

BIDIRECTIONAL BIT #	FUNCTION	ARINC BIT #
15, 14		13, 12
13	LSB	11
12, 11	SDI or Data	10, 9
10, 9	SSM Status	31, 30
8	Parity	32
7 - 00	Label	1 - 8

TABLE 3 - WORD 2 FORMAT

BIDIRECTIONAL BIT #	FUNCTION	ARINC BIT #
15	Sign	29
14	MSB	28
13 - 00	Data	27 - 14

If the receiver input data word string is broken before the entire data word is received, the receiver will reset and ignore the partially received data word.

If the transmitter is used to transmit consecutive data words, each word will be separated by a four (4) bit "null" state (both positive and negative outputs will maintain a zero (0) volt level).

### RECEIVER OPERATION

Since the two receivers are functionally identical, only one will be discussed in detail, and the block diagram will be used for reference in this discussion. The receiver consists of the following circuits:

- The Line Receiver functions as a voltage level translator. It transforms the 10 volt differential line voltage, ARINC 429 format, into 5 volt internal logic level.
- The output of the Line Receiver is one of two inputs to the Self-Test Data Selector (SEL). The other input to the Data Selector is the Self-Test Signal from the Transmitter section.
- The incoming data, either Self-Test or ARINC 429, is double sampled by the Word Gap Timer

to generate a Data Clock. The Receiver sample frequency (RCVCLK), 1 MHz, or 125 KHz, is generated by the Receiver/Transmitter Timing Circuit. This sampling frequency is ten times the Data Rate to ensure no data ambiguity.

- The derived data clock then shifts the data down a 32-bit long Data Shift Register (Data S/R1). The Data Word Length is selectable for either 25-Bits or 32-Bits long by the Control Signal (WLSEL). As soon as the data word is completely received, an internal signal (WDCNT1) is generated by the Word Gap Timer Circuit.
- The Source/Destination (S/D) Decoder compares the user set code (X and Y) with Bits 9 and 10 of the Data Word. If the two codes are matched, a positive signal is generated to enable the WDCNT1 signal to latch in the received data. Otherwise, the data word is ignored and no latching action takes place. The S/D Decoder can be Enabled and Disabled by the control signal S/D ENB. If the data word is latched, an indicator flag (D/R1) is set. This indicates a valid data word is ready to be fetched by the user.
- The parity of the incoming word is checked and the status (i.e., logic "0" for odd parity and logic "1" for even parity) stored in the receiver latch and outputted on BD08 during the Word No. 1.
- Assuming the user desires to access the data, he first sets the Data Select Line (SEL) to a Logic "0" level and pulses the Enable (EN1) line. This action causes the Data Selector (SEL1) to select the first-data word, which contains the label field and Enable it onto the Data Bus. To obtain the second data word, the user sets the SEL line to a Logic "1" level and pulse the Enable (EN1) line again. The Enable pulse duration is matched to the user circuit requirement needed to latch in the Data Word from the Data Bus. The second Enable pulse is also used to reset the Device Ready (D/R1) flip-flop. This completes a receiving cycle.

### TRANSMITTER OPERATION

The Transmitter section consists of an 8-word deep by 31-Bit long FIFO Memory, Parity Generator, Transmitter Word Gap Timing Circuit and Driver Circuit.

- The FIFO Memory is organized in such a way that data loaded in the input register is automatically transferred to the output register for Serial Data Transmission. This eliminates a large amount of data managing time since the data need not be clocked from the input register to the output register. The FIFO input register is made up of two sets of 16 D-type flip-flops, which are clocked by the two parallel load signals (PL1 and PL2). The data from the Data Bus is clocked into the D-type flip-flop on the positive going edge of the PL signals. If the FIFO memory is initially empty, or the stack is not full, the data will be automatically transferred down the Memory Stack and into the output register or to the last empty FIFO storage register. If the



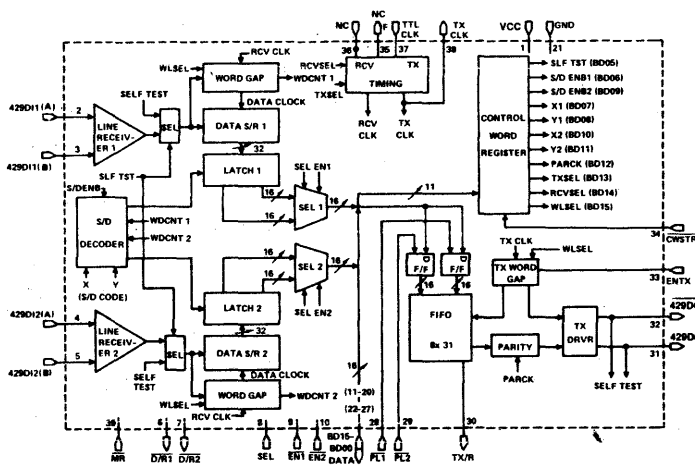
Transmitter Enable signal (ENTX) is not active, a Logic "0", the data remains at the output register. The FIFO Memory has storage locations to hold eight 31-Bit words. If the memory is full and new data is again strobed with  $\overline{PL}$ , the old data at the input register is written over by the new data. Data will remain in the Memory until ENTX goes to a Logic "1". This activates the FIFO Clock and data is shifted out serially to the Transmitter Driver.

- The Output Register of the FIFO is designed such that it can shift out a word of 25-Bits long or 32-Bits long. This word length is again controlled by the WLSEL bit. The TX Word Gap Timer Circuit also automatically inserts a gap equivalent to 4-Bit Times between each word. This gives a minimum requirement of 29-Bit time or 36-Bit time for each word transmission. Assuming the signal, ENTX, remains at a Logic "1", a transfer to stack signal is generated to transfer the data down the Memory Stack one position. This action is continued until the last word is shifted out of the FIFO memory. At this time a Transmitter Ready (TX/R) flag is generated to signal the user that the Transmitter is ready to receive eight more data words.
- A Bit Counter is used to detect the last Bit shifted out of the FIFO memory and replaces it with the Parity Bit generated by the Parity Generator. The Parity Generator has a control signal, Parity Check (PARCK), which establishes whether odd or even parity is inserted into the output data word. PARCK set to a logic "0" will result in odd parity and when set to a logic "1" will result in even parity.

Figure 2 shows the typical interface timing control of the ARINC Chip for Receiving function and for Transmitting function. Timing sequence for loading the Transmitter FIFO Memory is shown in Timing Interval A. A Transmitter Ready (TX/R) Flag signals the user that the Transmitter Memory is empty. The user then Enables the Transmitter Data, a 16-Bit word, on the Data Bus and strobes the Transmitter with a Parallel Load ( $\overline{PL1}$ ) Signal. The second part of the 32-Bit word is similarly loaded into the Transmitter with  $\overline{PL2}$ , which also initiates data transfer to stack. This is continuous until the Memory is full, which is eight 31-Bit words. The user must keep track of the number of words loaded into the Memory to ensure no data is written over by other data. During the time the user is loading the Transmitter, he does not have to service the Receiver, even if the Receiver flags the user with the signal  $\overline{D/R1}$  that a valid received word is ready to be fetched. This is shown by the Timing Interval B. If the user decides to obtain the received data before the Transmitter is completely loaded, he sets the two parallel load signals ( $\overline{PL1}$  and  $\overline{PL2}$ ) at a Logic "1" state, and strobe  $\overline{EN1}$  while the signal SEL is at a Logic "0" state. After the negative edge of  $\overline{EN1}$ , the first 16-Bit segment of the received word becomes valid on the Data Bus. At the positive edge of  $\overline{EN1}$ , the user should toggle the signal SEL to ready the Receiver for the second 16-Bit word. Strobing the Receiver with  $\overline{EN1}$ , the second time, enables the second 16-Bit word and resets the Receiver Ready Flag  $\overline{D/R1}$ . The user should now reset the signal SEL to a Logic "0" state to ready the Receiver for another Read Cycle. During the time period that the user is fetching the received words, he can load the transmitter. This is done by interlacing the  $\overline{PL}$  signals with the  $\overline{EN}$  signals as shown in the Timing Interval B. Servicing the Receiver 2 is similar and is illustrated by Timing Interval C. Timing Interval D shows the rest of the Transmitter loading sequence and the beginning of the transmission by switching the signal TX Enable to a Logic "1" state. Timing Interval E is the time it takes to transmit all data from the FIFO Memory, either 288 Bit times or 232 Bit times.

**SAMPLE INTERFACE TECHNIQUE**

From Figure 1, one can see that the Data Bus is time shared between the Receiver and the Transmitter. Therefore, bus controlling must be synchronously shared between the Receiver and the Transmitter.



**FIGURE 1. SINGLE CHIP ARINC 429 INTERFACE FUNCTIONAL BLOCK DIAGRAM.**

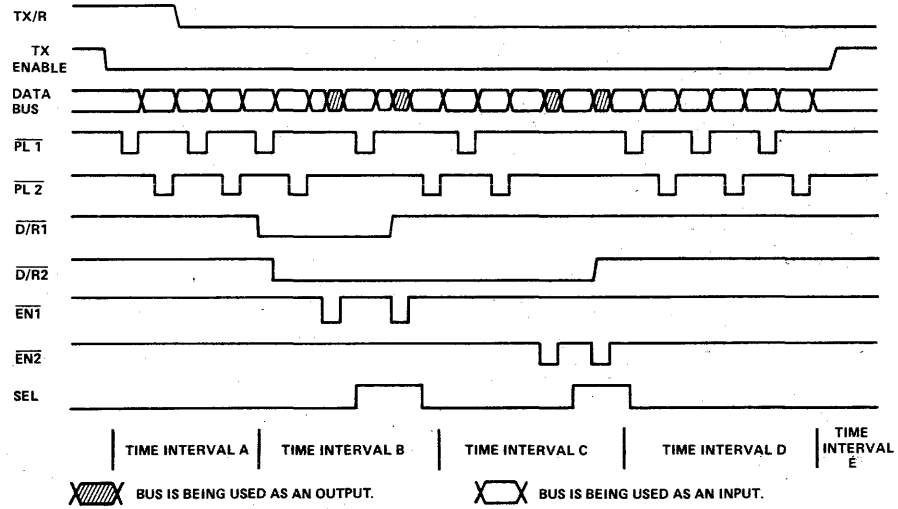


FIGURE 2. TYPICAL INTERFACE TIMING SEQUENCE

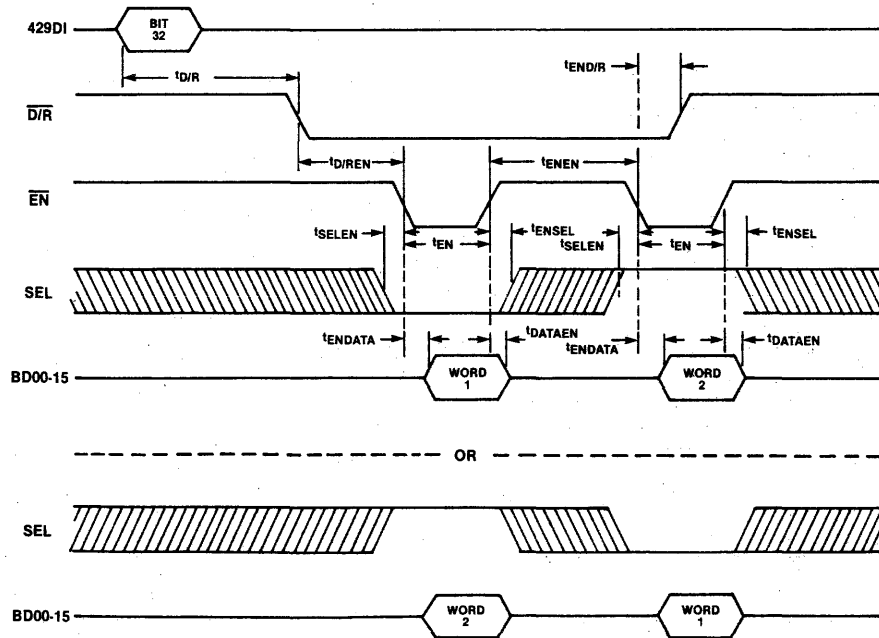


FIGURE 3 RECEIVER TIMING

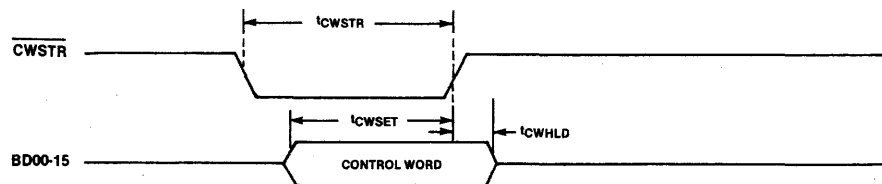


FIGURE 4 CONTROL WORD TIMING

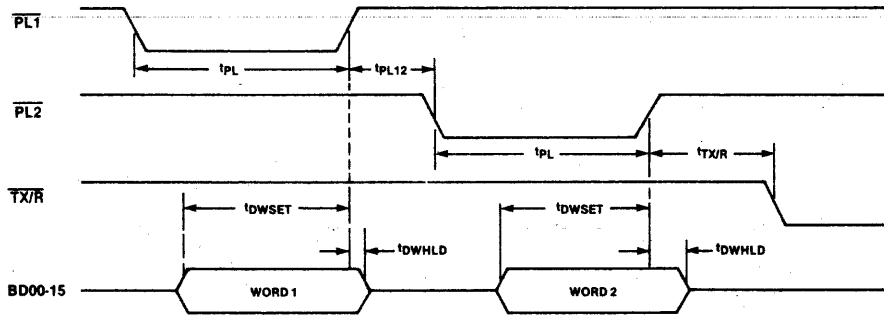


FIGURE 5 TRANSMITTER FIFO WRITE TIMING

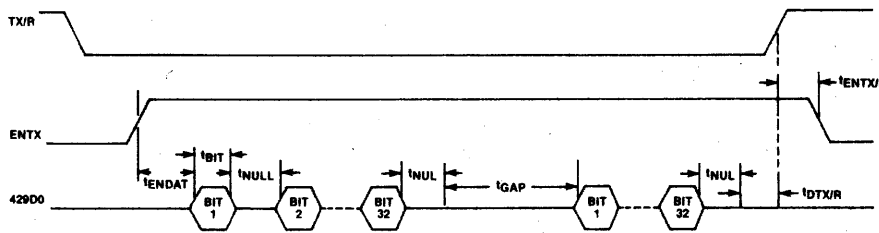


FIGURE 6 TRANSMITTER OUTPUT TIMING

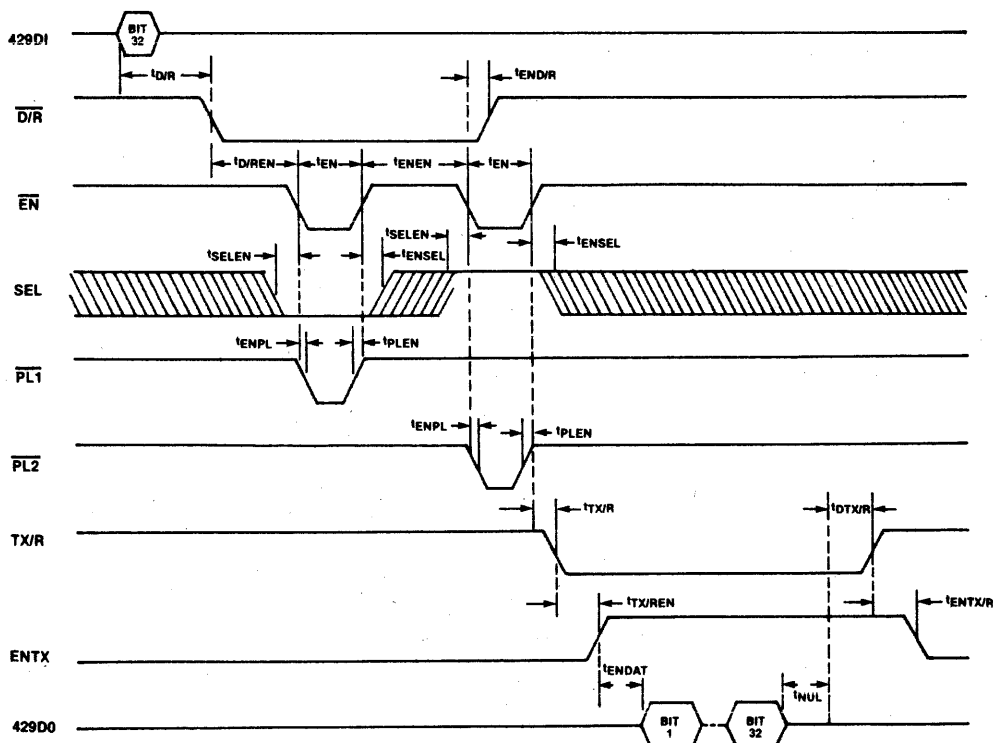
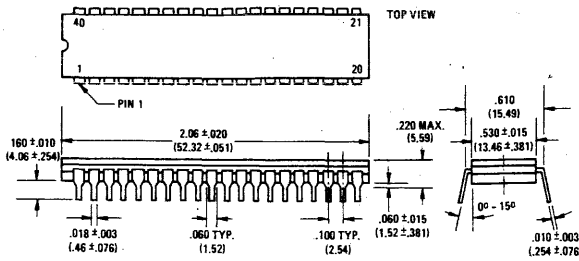


FIGURE 7 REPEATER OPERATION TIMING

Harris Semiconductor  
CUSTOM/SEMICUSTOM

## Packaging

40-PIN CER-DIP



## Ordering Information

**HS** 1-3282-8

HARRIS CUSTOM INTEGRATED CIRCUITS DIVISION

TEMPERATURE:  
8: Mil. Std. 883B  
5: 0°C to +70°C

DEVICE NUMBER

PACKAGE CODE  
1: Ceramic

- All dimensions in inches; millimeters are shown in parentheses.
- All dimensions ± .010 (± 0.25mm) unless otherwise shown.

## Test Product Flow

### HARRIS SEMICONDUCTOR PRODUCT FLOW MIL-STD-883, METHOD 5004 CLASS B 100% SCREENING PROCEDURE

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

Notes: Traceability.

All devices are assigned date code identification that provides traceability back to the inspection lot.

Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.

Branding.

All devices are branded with the part number and EIA date code.

Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.

## Sales Offices

### UNITED STATES

#### EASTERN REGION

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
TWX: 710-332-1074

106 Seventh Street  
Garden City, NY 11530  
(516) 747-6776

#### WESTERN REGION

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Suite 320  
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(714) 957-6557  
TWX: 910-595-1533

625 Ellis, Suite 300  
Mt. View, CA 94043  
(415) 964-6443  
TWX: 910-379-6431

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Minneapolis, MN 55420  
(612) 854-3224  
TWX: 910-576-3418

#### HOME OFFICE

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



# HARRIS

CUSTOM INTEGRATED CIRCUITS DIVISION



Preliminary

## Video Character Generator

### Features

- OPERATION (DOT CLOCK) FROM 1.0 TO 27 MHz
- DESIGNED TO INTERFACE WITH INTEL 8275 PROGRAMMABLE CRT CONTROLLER
- DIRECTLY DECODES (FROM ON-CHIP ROM);
  - ASCII 96 CHARACTER SET
  - 32 PSEUDO-GRAPHIC CHARACTERS
  - 10 OVERLAY PATTERNS
- EXPANDABLE CHARACTER TABLE
- 9 X 12 DOT MATRIX WITH DESCENDER CAPABILITY
- HANDLES VIDEO MODIFIERS;
  - BLANK
  - VIDEO SUPPRESS
  - REVERSE VIDEO
  - LIGHT ENABLE

### Pinout

WR	1	40	VCC
CS	2	39	GRAPHPE
A0	3	38	E8
S1	4	37	E7
S2	5	36	E6
S3	6	35	E5
LC3	7	34	E4
LC2	8	33	E3
LC1	9	32	E2
LC0	10	31	E1
EXEN	11	30	E0
ROMDIS	12	29	DOTCLK
CC6	13	28	RESET
CC5	14	27	CCLK
CC4	15	26	VSP
CC3	16	25	LTEN
CC2	17	24	RVV
CC1	18	23	BLK
CC0	19	22	VID2
GND	20	21	VID1

LC0-3 - LINE COUNT  
 CC0-6 - CHARACTER CODE  
 ROMDIS - ROM DISABLE  
 LTEN - LIGHT ENABLE  
 RVV - REVERSE VIDEO  
 VSP - VIDEO SUPPRESS  
 BLK - BLANK  
 S1-3 - SPECIAL FUNCTION  
 EXEN - EXPANSION ENABLE  
 E0-E8 - EXPANSION INPUTS  
 GRAPHPE - GRAPHICS ENABLE  
 WR - WRITE  
 CS - CHIP SELECT  
 A0 - ADDRESS 0  
 VID1-2 - VIDEO OUTPUTS  
 RESET - RESET  
 DOTCLK - DOT CLOCK  
 CCLK - CHARACTER CLOCK

### Description

The HS-3819 is a CMOS/LSI Video Character Generator designed to help interface an Intel 8275 Programmable CRT Controller to a video monitor. The character generator must be supplied with a clock frequency of between 1 and 27 MHz which will be used as the dot clock. This signal is then divided by nine to form the character clock output needed by the CRT Controller. The HS-3819 then converts character data into a video output signal, through use of an internal (ROM) character table. Stored in this ROM are the standard 96 ASCII characters, 32 pseudo-graphic characters and 10 overlay patterns used to modify characters. Additional characters, if needed, can easily be decoded from an external memory field.

### Standard Character Set

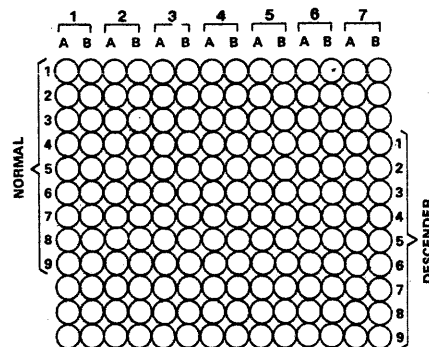
	0	1	2	3	4	5	6	7
0	NULL	Ⓢ	SP	0	Ⓟ	P	~	p
1	EM	Ⓢ	!	1	A	Q	a	q
2	!	Ⓢ	"	2	B	R	b	r
3	▶	Ⓢ	#	3	C	S	c	s
4	■	Ⓢ	\$	4	D	T	d	t
5	?	Ⓢ	%	5	E	U	e	u
6	▣	Ⓢ	&	6	F	V	f	v
7	▢	Ⓢ	'	7	G	W	g	w
8	▣	Ⓢ	(	8	H	X	h	x
9	▣	Ⓢ	)	9	I	Y	i	y
A	Ⓢ	Ⓢ	*	:	J	Z	j	z
B	Ⓢ	Ⓢ	+	;	K	[	k	{
C	Ⓢ	Ⓢ	<	L	\			
D	Ⓢ	Ⓢ	=	M	]	m	}	}
E	Ⓢ	Ⓢ	>	N	^	n	~	~
F	Ⓢ	Ⓢ	?	O	_	o	ƒ	ƒ

### Overlay Patterns

CRISS CROSS	Ⓢ
UNDERLINE	—
CRISS CROSS	Ⓢ
DIAGONAL STRIKE	Ⓢ
DIAGONAL STRIKE	Ⓢ
DASHED UNDERLINE	Ⓢ
DASHED UNDERLINE	Ⓢ
DASHED UNDERLINE	Ⓢ
OPEN BOX	□
OPEN BOX	□
UNDERDOT	·
UNDERDOT	·
REVERSE DIAGONAL	Ⓢ
REVERSE DIAGONAL	Ⓢ
DOUBLE UNDERLINE	Ⓢ

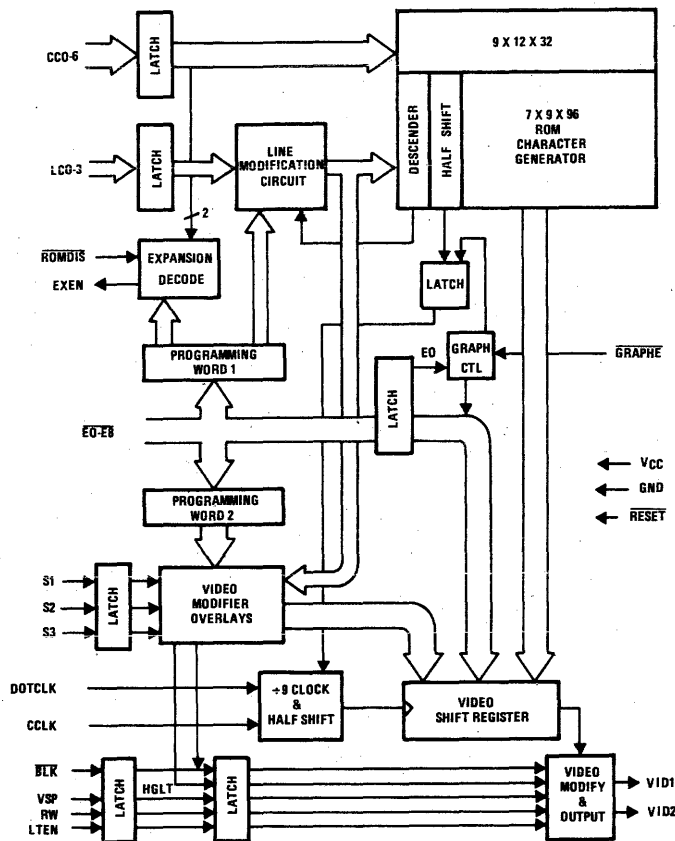
CRISS CROSS HATCH  
 HORIZONTAL STRIKE THROUGH

### Dot Matrix

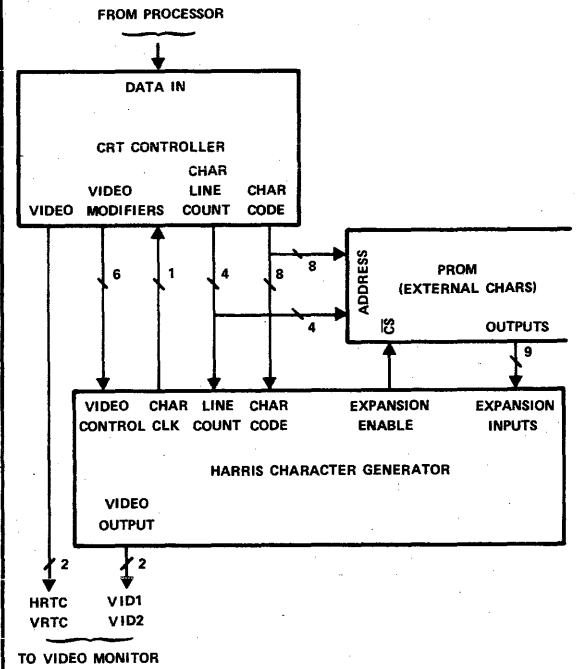


- NOTES:
1. EACH ROW MAY HAVE DOTS IN SET A OR SET B ONLY.
  2. EACH CHARACTER MAY HAVE DOTS IN NORMAL SET OR DESCENDER SET ONLY.

## HS-3819 Block Diagram



## System Application



### Sales Offices

#### UNITED STATES

##### EASTERN REGION

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
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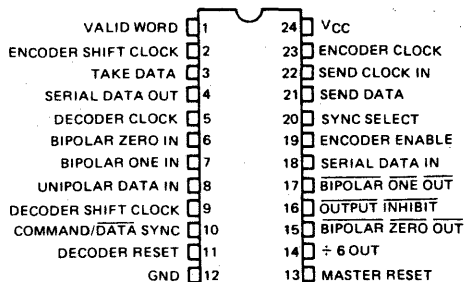
# HARRIS

SEMICONDUCTOR CUSTOM INTEGRATED CIRCUITS DIVISION

### Features

- SUPPORT OF MIL-STD-1553
- 1.0 MEGABIT/SEC DATA RATE
- SYNC IDENTIFICATION AND LOCK-IN
- CLOCK RECOVERY
- MANCHESTER II ENCODE, DECODE
- SEPARATE ENCODE AND DECODE
- LOW OPERATING POWER: 50mW AT 5 VOLTS
- FULL MILITARY TEMPERATURE RANGE
- FUNCTIONAL TOTAL DOSE ...  $1 \times 10^4$  RAD (Si)
- LATCH-UP FREE TO ...  $5 \times 10^{11}$  RAD (Si)/sec

### Pinout



### Description

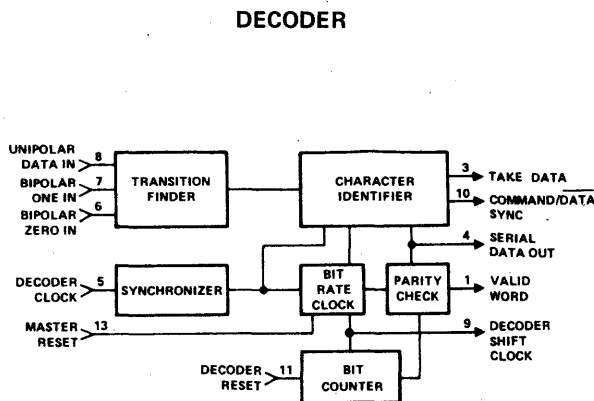
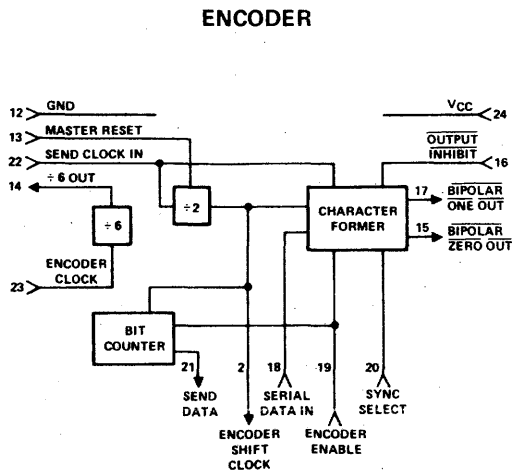
The Harris HS-15530RH is a high performance, radiation resistant, CMOS device intended to service the requirements of MIL-STD-1553 and similar Manchester II encoded, time division multiplexed serial data protocols. This LSI chip is divided into two sections, an Encoder and a Decoder. These sections operate completely independent of each other, except for the Master Reset function.

This circuit provides many of the requirements of MIL-STD-1553. The Encoder produces the sync

pulse and the parity bit as well as the encoding of the data bits. The Decoder recognizes the sync pulse and identifies it as well as decoding the data bits and checking parity.

This integrated circuit is fully guaranteed to support the 1MHz data rate of MIL-STD-1553 over both temperature and voltage while residing in a radiation environment. It interfaces with CMOS, TTL or N channel support circuitry, and uses a standard 5 volt supply.

### Block Diagrams



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage	+7.0V
Input or Output Voltage Applied	GND -0.3V to VCC +0.3V
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	-55°C to +125°C

**ELECTRICAL CHARACTERISTICS** VCC = 5.0V ±5% TA = Operating Temperature Range

SYMBOL	PARAMETER	MINIMUM	TYPICAL	MAXIMUM	UNITS	TEST CONDITIONS	
D.C.	VIH	Logical "1" Input Voltage	70% VCC		V	0V ≤ VIN ≤ VCC IOH = -3mA IOL = 1.8mA VIN = VCC = 5.25V Outputs Open VCC = 5.25V, f = 1MHz	
	VIL	Logical "0" Input Voltage		20% VCC	V		
	VIHC	Logical "1" Input Voltage (Clock)	VCC -0.5		V		
	VILC	Logical "0" Input Voltage (Clock)		GND +0.5	V		
	IIL	Input Leakage	-1.0		μA		
	VOH	Logical "1" Output Voltage	2.4		V		
	VOL	Logical "0" Output Voltage		0.4	V		
	ICCSB	Supply Current Standby		0.5	2		mA
	ICCOP	Supply Current Operating*		8.0	10.0		mA
	CIN	Input Capacitance*		5.0	7.0		pF
CO	Output Capacitance*		8.0	10.0	pF		

\*Guaranteed and sampled but not 100% tested.

**ENCODER TIMING** VCC = 5.0V ±5% TA = Operating Temperature Range

A.C.	FEC	Encoder Clock Frequency			13	MHz	CL = 50pF
	FESC	Send Clock Frequency			2.16	MHz	
	TECR	Encoder Clock Rise Time			8	ns	
	TECF	Encoder Clock Fall Time			8	ns	
	FED	Data Rate			1.08	MHz	
	TMR	Master Reset Pulse Width	150			ns	
	TE1	Shift Clock Delay			125	ns	
	TE2	Serial Data Setup	75			ns	
	TE3	Serial Data Hold	75			ns	
	TE4	Enable Setup	90			ns	
TE5	Enable Pulse Width	80			ns		
TE6	Sync Setup	55			ns		
TE7	Sync Pulse Width	150			ns		
TE8	Send Data Delay	-10		50	ns		
TE9	Bipolar Output Delay			130	ns		

**DECODER TIMING** VCC = 5.0V ±5% TA = Operating Temperature Range

A.C.	FDC	Decoder Clock Frequency			13	MHz	CL = 50pF
	TDCR	Decoder Clock Rise Time			8	ns	
	TDCF	Decoder Clock Fall Time			8	ns	
	FDD	Data Rate			1.08	MHz	
	TDR	Decoder Reset Pulse Width	150			ns	
	TDRS	Decoder Reset Setup Time	75			ns	
	TMR	Master Reset Pulse Width	150			ns	
	TD1	Bipolar Data Pulse Width	TDC +10			ns	
	TD2	Sync Transition Span		18TDC		ns	
	TD3	One Zero Overlap			TDC -10	ns	
	TD4	Short Data Transition Span		6TDC		ns	
TD5	Long Data Transition Span		12TDC		ns		
TD6	Sync Delay (ON)		40	110	ns		
TD7	Take Data Delay (ON)		50	110	ns		
TD8	Serial Data Out Delay		80	80	ns		
TD9	Sync Delay (OFF)		90	110	ns		
TD10	Take Data Delay (OFF)		110	110	ns		
TD11	Valid Word Delay		90	110	ns		

NOTE ① : 15TDC +10 = [15 (Decoder Clock Period)] +10ns TDC = Decoder Clock Period =  $\frac{1}{FDC}$   
 These parameters are guaranteed but not 100% tested.

## Pin Assignments

PIN	SECTION	NAME	DESCRIPTION
1	Decoder	VALID WORD	Output high indicates receipt of a valid word.
2	Encoder	ENCODER SHIFT CLOCK	Output for shifting data into the Encoder. This clock shifts data on a low-to-high transition.
3	Decoder	TAKE DATA	Output is high during receipt of data after identification of a sync pulse.
4	Decoder	SERIAL DATA OUT	Delivers received data in correct NRZ format.
5	Decoder	DECODER CLOCK	Input drives the transition finder, and the synchronizer which in turn supplies the clock to the balance of the Decoder.
6	Decoder	BIPOLAR ZERO IN	A high input should be applied when the bus is in its negative state. This pin must be held high when the Unipolar input is used.
7	Decoder	BIPOLAR ONE IN	A high input should be applied when the bus is in its positive state, this pin must be held low when the Unipolar input is used.
8	Decoder	UNIPOLAR DATA IN	With pin 6 high and pin 7 low, this pin enters unipolar data into the transition finder circuit. If not used this input must be held low.
9	Decoder	DECODER SHIFT CLOCK	Output which delivers a frequency (Decoder Clock $\div$ 12), synchronized by the recovered serial data stream.
10	Decoder	COMMAND SYNC	Output of a high from this pin occurs during output of decoded data which was preceded by a Command (or Status) synchronizing character. A low output indicates a Data synchronizing character.
11	Decoder	DECODER RESET	A high input to this pin during a rising edge of DECODER SHIFT CLOCK resets the decoder bit counting logic to a condition ready for a new word.
12	Both	GROUND	Ground supply pin.
13	Both	MASTER RESET	A high on this pin clears 2:1 counters in both the Encoder and Decoder.
14	Encoder	$\div$ 6 OUT	Output from 6:1 divider which is driven by the ENCODER CLOCK.
15	Encoder	<u>BIPOLAR ZERO OUT</u>	An active low output designed to drive the zero or negative sense of a bipolar line driver.
16	Encoder	<u>OUTPUT INHIBIT</u>	A low on this input forces pin 15 and pin 17 high, the inactive states.
17	Encoder	<u>BIPOLAR ONE OUT</u>	An active low output designed to drive the one or positive sense of a bipolar line driver.
18	Encoder	SERIAL DATA IN	Accepts a serial data stream at a data rate equal to ENCODER SHIFT CLOCK.
19	Encoder	ENCODER ENABLE	A high on this input initiates the encode cycle. (Subject to the preceding cycle being complete.)
20	Encoder	SYNC SELECT	Actuates command sync for an input high and data sync for an input low.
21	Encoder	SEND DATA	Is an active high output which enables the external source of serial data.
22	Encoder	SEND CLOCK IN	Clock input at a frequency equal to the data rate X2.
23	Encoder	ENCODER CLOCK	Input to the 6:1 divider.
24	Both	VCC	Positive supply pin.



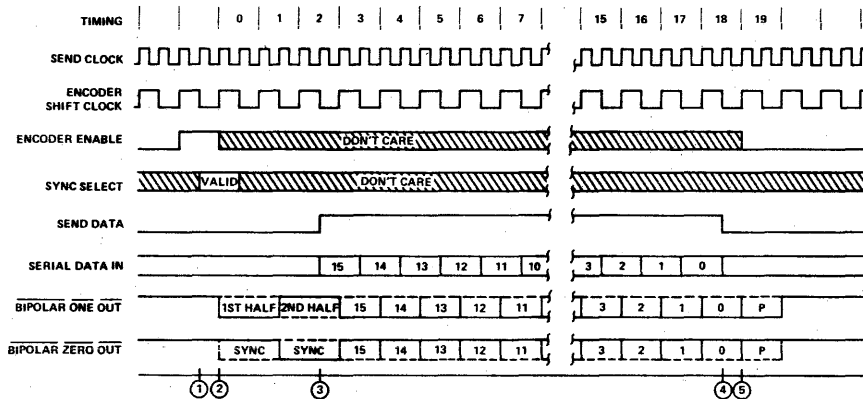
### Encoder Operation

The Encoder requires a single clock with a frequency of twice the desired data rate applied at the SEND CLOCK input. An auxiliary divide by six counter is provided on chip which can be utilized to produce the SEND CLOCK by dividing the DECODER CLOCK.

The Encoder's cycle begins when ENCODER ENABLE is high during a falling edge of ENCODER SHIFT CLOCK (1). This cycle lasts for one word length or twenty ENCODER SHIFT CLOCK periods. At the next low-to-high transition of the ENCODER SHIFT CLOCK, a high at SYNC SELECT input actuates a command sync or a low will produce a data sync for that word (2). When the Encoder is ready to accept data, the SEND DATA output will go high and remain high for sixteen ENCODER SHIFT CLOCK periods (3). During these sixteen periods the data should be

clocked into the SERIAL DATA input with every low-to-high transition of the ENCODER SHIFT CLOCK (3) - (4). After the sync and the Manchester II coded data are transmitted through the BIPOLAR ONE and BIPOLAR ZERO outputs, the Encoder adds on an additional bit which is the parity for that word (5). At any time a low on OUTPUT INHIBIT input will force both bipolar outputs to a high state but will not affect the Encoder in any other way.

To abort the Encoder transmission a positive pulse must be applied at MASTER RESET. Anytime after or during this pulse, a low to high transition on SEND CLOCK clears the internal counters and initializes the Encoder for a new word.



### Decoder Operation

The Decoder requires a single clock with a frequency of 12 times the desired data rate applied at the DECODER CLOCK input. The Manchester II coded data can be presented to the Decoder in one of two ways. The BIPOLAR ONE and BIPOLAR ZERO inputs will accept data from a comparator sensed transformer coupled bus as specified in Military Spec 1553. The UNIPOLAR DATA input can only accept non-inverted Manchester II coded data. (e.g. from BIPOLAR ZERO OUT of an Encoder.)

The Decoder is free running and continuously monitors its data input lines for a valid sync character and two valid Manchester data bits to start an output cycle. When a valid sync is recognized (1), the type of sync is indicated on COMMAND/DATA SYNC output. If the sync character was a command sync, this output will go high (2) and remain high for sixteen DECODER SHIFT CLOCK periods (3), otherwise it will remain low. The TAKE DATA output will go high and remain high (2) - (3) while the

Decoder is transmitting the decoded data through SERIAL DATA OUT. The decoded data available at SERIAL DATA OUT is in a NRZ format. The DECODER SHIFT CLOCK is provided so that the decoded bits can get shifted into an external register on every low-to-high transition of this clock (2) - (3).

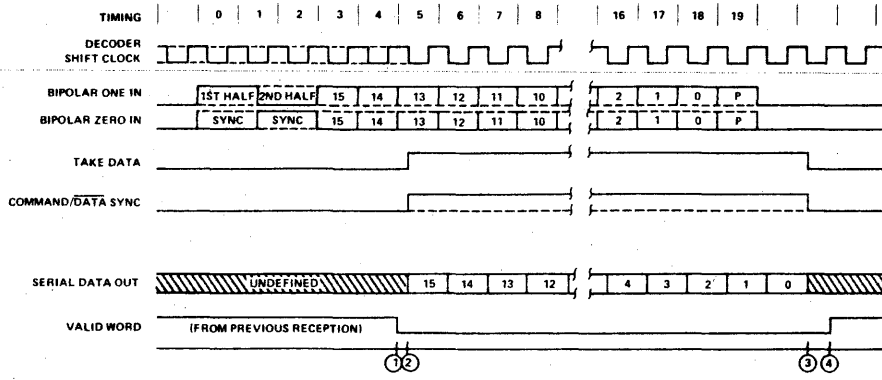
After all sixteen decoded bits have been transmitted (3) the data is checked for odd parity. A high on VALID WORD output (4) indicates a successful reception of a word without any Manchester or parity errors. At this time the Decoder is looking for a new sync character to start another output sequence.

At any time in the above sequence a high input on DECODER RESET during a low-to-high transition of DECODER SHIFT CLOCK will abort transmission and initialize the Decoder to start looking for a new sync character.

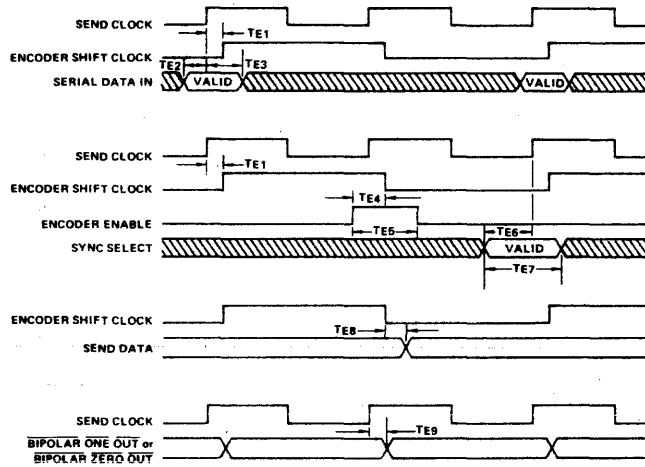


# Operational Description (continued)

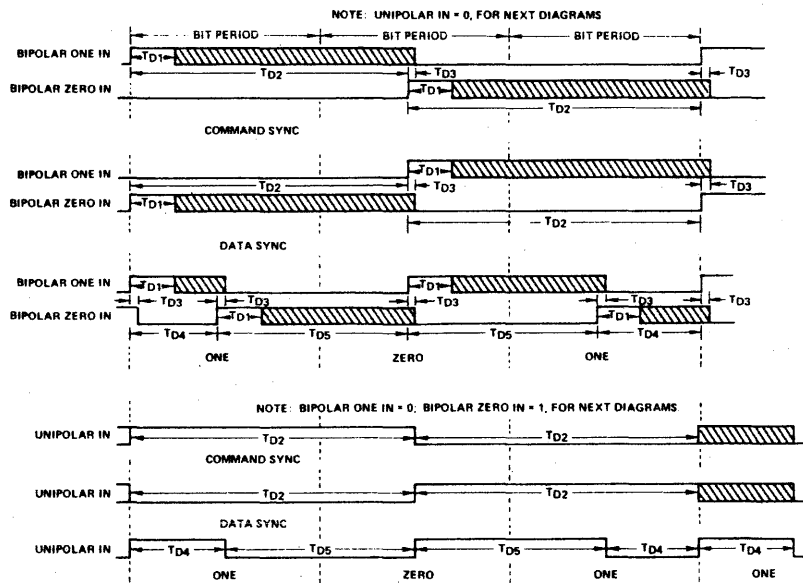
## Decoder Operation (continued)



## Encoder Timing



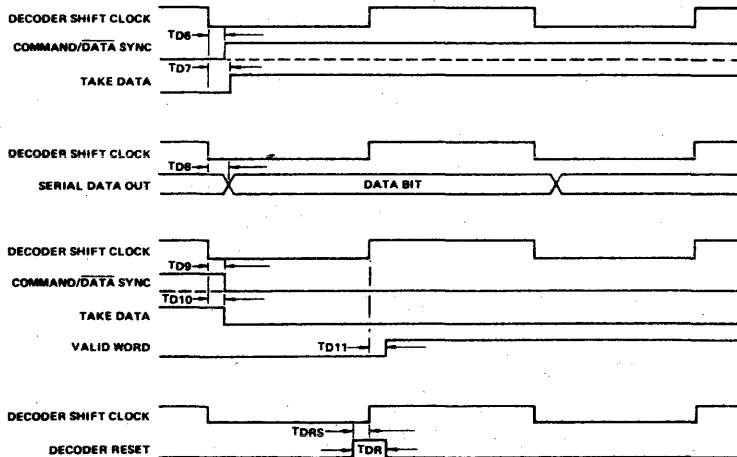
## Decoder Timing



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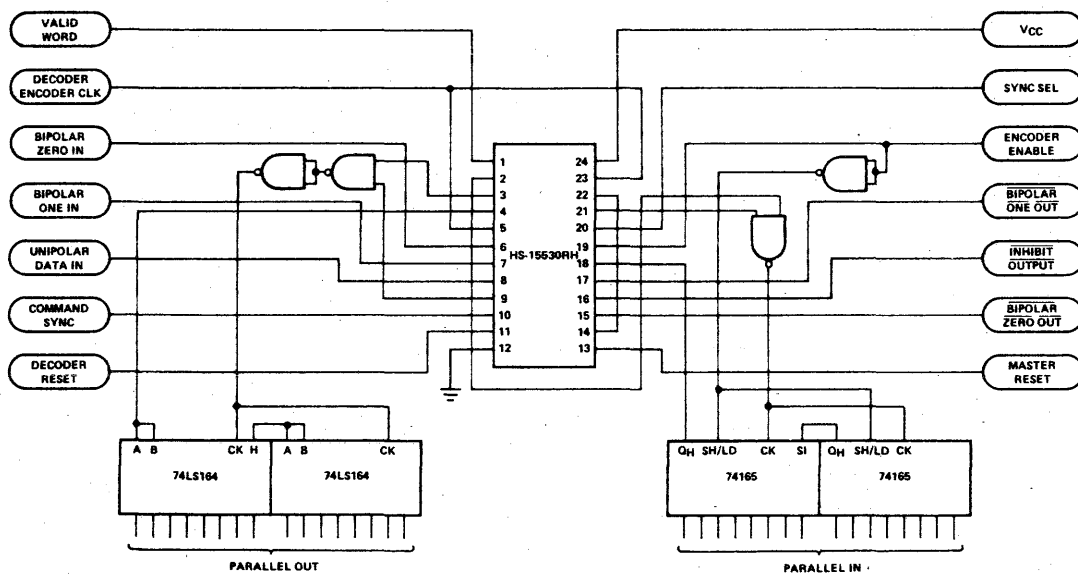


Decoder Timing (continued)

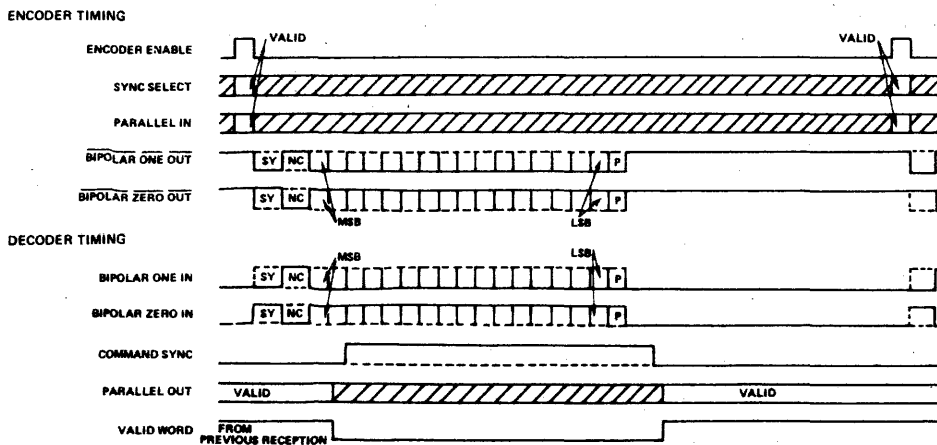


Applications

How to Make Our MTU Look Like a Manchester Encoded UART



Typical Timing Diagrams for a Manchester Encoded UART







The 1553 standard defines a time division multiplexed data bus for application within aircraft. The bus is defined to be bipolar, and encoded in a Manchester II format, so no DC component appears on the bus. This allows transformer coupling and excellent isolation among systems and their environment.

The HS-15530RH supports the full bipolar configuration, assuming a bus driver configuration similar to that in Figure 1. Bipolar inputs from the bus, like Figure 2, are also accommodated.

The signaling format in MIL-STD-1553 is specified on the assumption that the network of 32 or fewer terminals are controlled by a central control unit by means of Command Words, and Data. Terminals respond with Status Words, and Data. Each word is preceded by a synchronizing pulse, and followed by parity bit, occupying a total of 20  $\mu$ sec. The word formats are shown in Figure 4. The special abbreviations are as follows:

- P Parity, which is defined to be odd, taken across all 17 bits.
- R/T Receive on logical zero, transmit on ONE.
- ME Message Error if logical 1.
- TF Terminal Flag, if set, calls for controller to request self-test data.

The paragraphs above are intended only to suggest the content of MIL-STD-1553, and do not completely describe its bus requirements, timing or protocols.

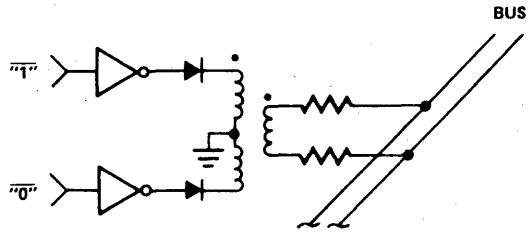


FIGURE 1 – Simplified MIL-STD-1553 Driver

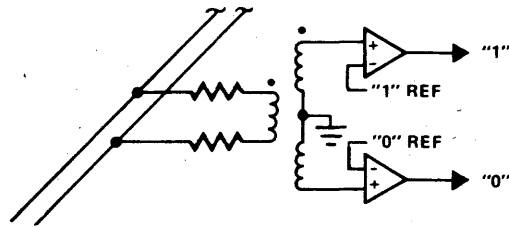


FIGURE 2 – Simplified MIL-STD-1553 Receiver

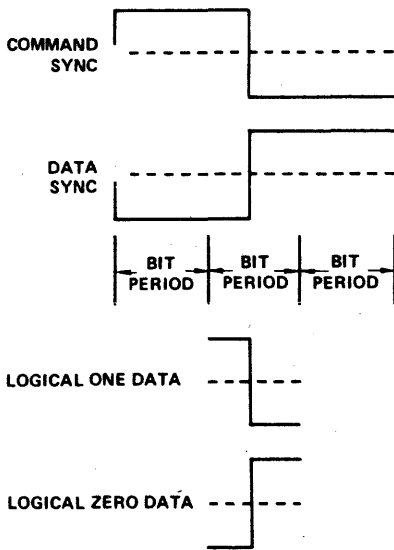


FIGURE 3 – MIL-STD-1553 Character Formats

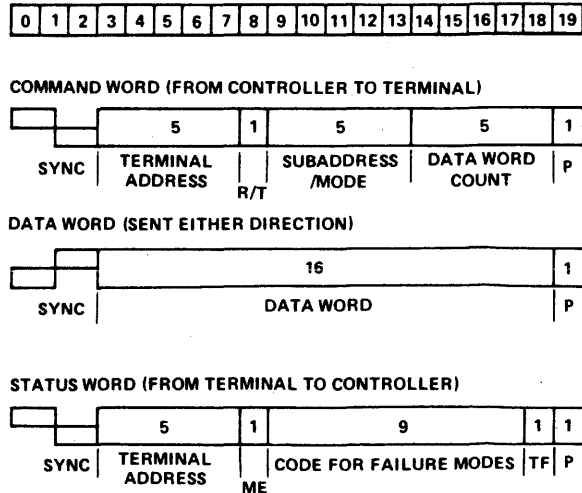


FIGURE 4 – MIL-STD-1553 Word Formats

NOTE: This page is a summary of MIL-STD-1553 and is not intended to describe the operation of the HS-15530RH

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## Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^4$  Rad(Si +10% from a Gamma Cell 220 Cobalt 60 source or equivalent. The samples shall be biased at 5 volts with all inputs high. The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4) ICCSB at  $V_{CC} = 5$  volts will be measured and recorded for each device within one hour ( $\pm 15$  minutes) after irradiation. The lot will be accepted only if the average of these measure values is  $\leq 5$ mA.

## Radiation Effects

The HS--15530RH is processed with the same mask set as is used for HARRIS' equivalent commercial part. Latchup free operation, achieved by the use of special starting material and improved total dose hardness, is obtained with special high temperature processing cycles. These process techniques can, in principal, be applied to any standard HARRIS CMOS product.

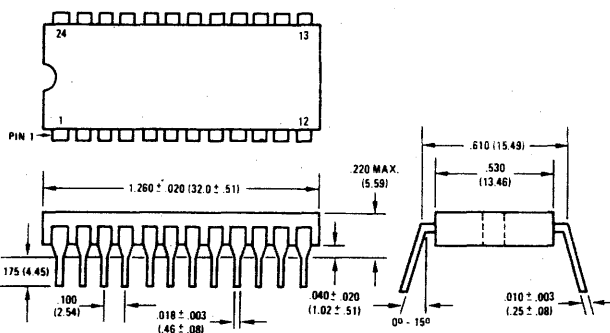
The primary failure mode under exposure to ionizing radiation is an increase in static leakage current (ICCSB). Functional failure due to the increased leakage currents will typically occur for dose levels in excess of  $5 \times 10^4$  RAD-Si. AC and DC parameters other than ICC will change less than 10% for total dose levels under  $5 \times 10^4$  RAD-Si. The excess leakage currents will anneal at room temperature and are typically reduced by a factor of 3-10 within 24 hours after irradiation.

On a production basis, HARRIS is able to perform screens only for a total dose hardness. Transient radiation tests, however, have shown the following results:

Latchup free doses  $\geq 5 \times 10^{11}$  rads/sec. Upset (loss of stored data) typically  $\geq 10^8$  rads/sec.

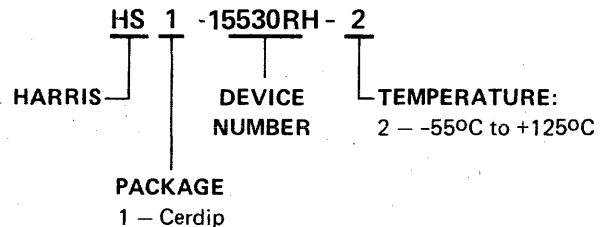
## Package

### 24 LEAD DIP



1. All dimensions in inches; millimeters are shown in parentheses.
2. All dimensions  $\pm .010$  ( $\pm 0.25$ mm) unless otherwise shown.

## Ordering Information



## Sales Offices

### UNITED STATES

#### EASTERN REGION

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
TWX: 710-332-1074

106 Seventh Street  
Garden City, NY 11530  
(516) 747-6776

#### WESTERN REGION

1503 South Coast Drive  
Suite 320  
Costa Mesa, CA 92626  
(714) 957-6557  
TWX: 910-595-1533

625 Ellis, Suite 300  
Mt. View, CA 94043  
(415) 964-6443  
TWX: 910-379-6431

#### MIDWEST REGION

2850 Metro Dr., Suite 703  
Minneapolis, MN 55420  
(612) 854-3224  
TWX: 910-576-3418

#### HOME OFFICE

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



# HARRIS

CUSTOM INTEGRATED CIRCUITS DIVISION



# HARRIS

## HS-6504RH

### 4096 x 1 CMOS RAM

### Preliminary

#### Features

- LOW POWER STANDBY . . . . . 100  $\mu$ W MAX.
- LOW POWER OPERATION . . . . . 25mW/MHz MAX.
- EXTREMELY LOW SPEED POWER PRODUCT
- FUNCTIONAL TOTAL DOSE . . . . .  $1 \times 10^5$  RAD Si
- DATA UPSET . . . . .  $>10^8$  RAD Si/SEC
- LATCH - UP FREE TO . . . . .  $>5 \times 10^{11}$  RAD Si/SEC
- TTL COMPATIBLE INPUT/OUTPUT
- THREE - STATE OUTPUT
- STANDARD JEDEC PINOUT
- FAST ACCESS TIME . . . . . 200nsec TYP
- MILITARY TEMPERATURE RANGE
- 18 PIN PACKAGE FOR HIGH DENSITY
- ON CHIP ADDRESS REGISTER

#### Description

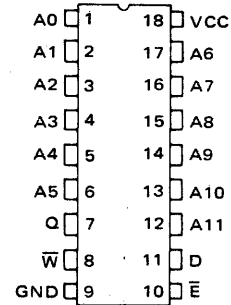
The HS-6504RH is a 4096 x 1 static CMOS RAM fabricated using the Harris Custom Integrated Circuits Division radiation hardened self-aligned silicon gate technology. The device utilizes synchronous circuitry to achieve high performance and low power operation.

On-chip latches are provided for addresses, data input and data output allowing efficient interfacing with microprocessor systems. The data output can be forced to a high impedance for use in expanded memory arrays.

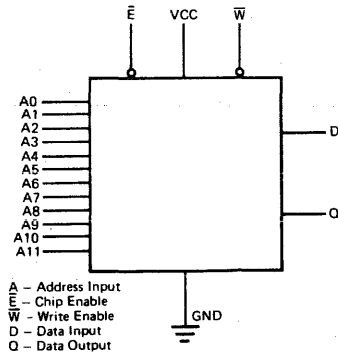
The HS-6504RH is a fully static RAM and may be maintained in any state for an indefinite period of time.

#### Pinout

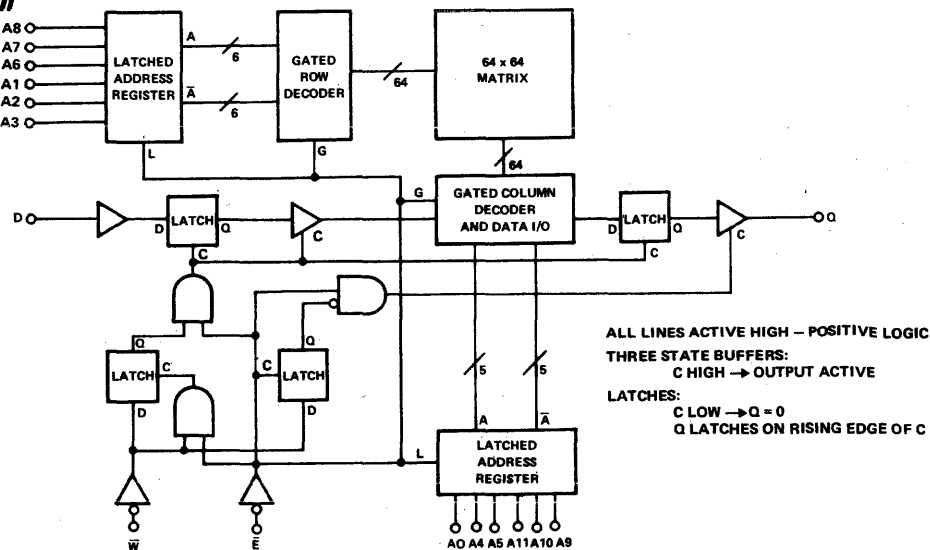
TOP VIEW



#### Logic Symbol



#### Functional Diagram



Information on this device is preliminary. Data is subject to change unless otherwise specifically agreed. No obligations are assumed for notice of change or future manufacture of this device.

CAUTION: These devices are sensitive to electrostatic discharge.

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## Symbols and Abbreviations

This data sheet utilizes a new set of specification nomenclature. This new format is an IEEE and JEDEC supported standard for semiconductor memories. It is intended to clarify the symbols, abbreviations and definitions, and to make all memory data sheets consistent. We believe that, once acclimated, you will find this standardized format easy to read and use.

### ELECTRICAL PARAMETER ABBREVIATIONS

All abbreviations use upper case letters with no subscripts. The initial symbol is one of these four characters:

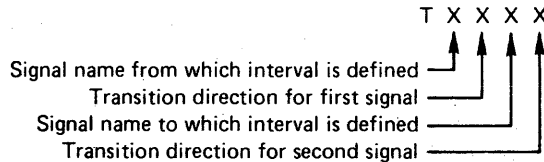
- V (Voltage)
- I (Current)
- P (Power)
- C (Capacitance)

The second letter specifies input (I) or output (O), and the third letter indicates the high (H), low (L) or off (Z) state of the pin during measurements. Examples:

- VIL — Input Low Voltage
- IOZ — Output Leakage Current

### TIMING PARAMETER ABBREVIATIONS

All timing abbreviations use upper case characters with no subscripts. The initial character is always T and is followed by four descriptors. These characters specify two signal points arranged in a "from-to" sequence that define a timing interval. The two descriptors for each signal point specify the signal name and the signal transitions. Thus the format is:



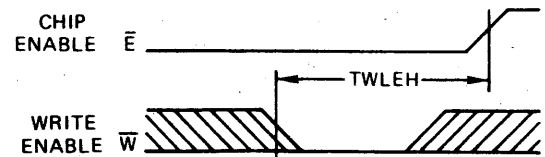
Signal Definitions:

- A = Address
- D = Data In
- Q = Data Out
- W = Write Enable
- E = Chip Enable
- S = Chip Select
- G = Output Enable

Transition Definitions:

- H = Transition to High
- L = Transition to Low
- V = Transition to Valid
- X = Transition to Invalid or Don't Care
- Z = Transition to Off (High Impedance)

EXAMPLE:



The example shows Write pulse setup time defined as TWLEH—Time from Write enable Low to chip Enable High.

### TIMING LIMITS

The table of timing values shows either a minimum or a maximum limit for each parameter. Input requirements are specified from the external system point of view. Thus, address set-up time is shown as a minimum since the system must supply at least that much time (even though most devices do not require it). On the other hand, responses from the memory are specified from the device point of view. Thus, the access time is shown as a maximum since the device never provides data later than that time.

### WAVEFORMS

WAVEFORM SYMBOL	INPUT	OUTPUT
	MUST BE VALID	WILL BE VALID
	CHANGE FROM H TO L	WILL CHANGE FROM H TO L
	CHANGE FROM L TO H	WILL CHANGE FROM L TO H
	DON'T CARE: ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
	—	HIGH IMPEDANCE



# Specifications HS-6504RH

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage - (VCC - GND)	-0.3V to +7.0V
Input or Output Voltage Applied	(GND -0.3V) to (VCC +0.3V)
Storage Temperature	-65°C to +150°C

## OPERATING RANGE

Operating Supply Voltage	4.5V to 5.5V
Operating Temperature	-55°C to +125°C

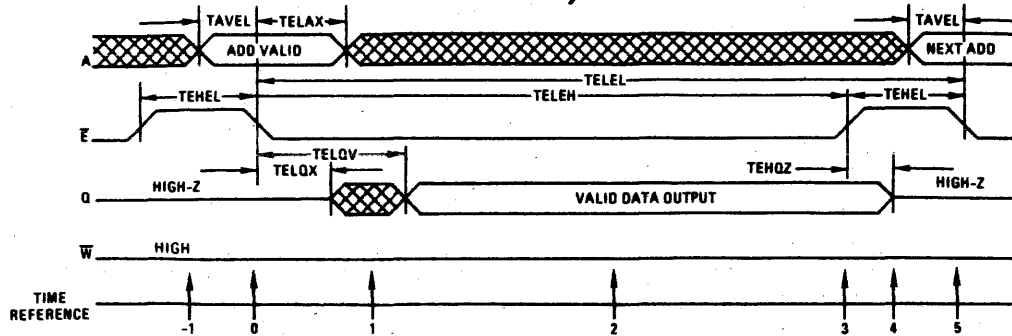
**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

## ELECTRICAL CHARACTERISTICS ⑤

SYMBOL	PARAMETER	TEMP. & VCC = OPERATING RANGE		TEMP = 25°C VCC = 5.0V TYPICAL ①		UNITS	TEST CONDITIONS	
		MIN	MAX	PRE RAD	POST RAD			
D.C.	ICCSB	Standby Supply Current		100	5	5	μA	IO = 0 VI = VCC or GND
	ICCOP	Operating Supply Current ②		7	4.5		mA	f = 1MHz, IO = 0 VI = VCC or GND
	ICCDR	Data Retention Current		50	3	3	μA	IO = 0, VCC = 3.0 VI = VCC or GND
	VCCDR	Data Retention Voltage		3	2.6	1.8	V	
	II	Input Leakage Current	-1	+1	0	0	μA	GND < VI < VCC
	IOZ	Output Leakage Current	-10	+10	±0.5	±0.5	μA	GND < VI < VCC
	VIL	Input Low Voltage	-0.3	0.8	1.5	1.3	V	
	VIH	Input High Voltage	VCC	VCC	2.7	2.3	V	
		All inputs except $\bar{E}$ , R/W	-2.0	+0.3				
	VIH	Input High Voltage for $\bar{E}$ , R/W	VCC	VCC	2.9	1.9	V	
			-1.5	+0.3				
	VOL	Output Low Voltage		0.4	0.2	0.15	V	IO = 2.0mA
VOH	Output High Voltage	2.4		4.6	4.0	V	IO = -1.0mA	
CI	Input Capacitance ③		8.0	5.0	5.0	pF	f = 1MHz VI = VCC or GND	
CO	Output Capacitance ③		10.0	6.0	6.0	pF	f = 1MHz VI = VCC or GND	
A.C.	TELOV	Chip Enable Access Time		300	150	180	ns	④
	TAVQV	Address Access Time		320	140	180	ns	④
	TELOX	Chip Enable Output Enable Time		100	50	60	ns	④
	TEHQZ	Chip Enable Output Disable Time		100	50	60	ns	④
	TELEH	Chip Enable Pulse Negative Width	300		150	180	ns	④
	TEHEL	Chip Enable Pulse Positive Width	120		40	80	ns	④
	TAVEL	Address Setup Time	20		-10	0	ns	④
	TELAX	Address Hold Time	50		30	35	ns	④
	TWLWH	Write Enable Pulse Width	80		20	30	ns	④
	TWLEH	Write Enable Pulse Setup Time	200		140	140	ns	④
	TWLEL	Early Write Pulse Setup Time	0		-15	-10	ns	④
	TWHEL	Write Enable Read Mode Setup Time	0		-15	-10	ns	④
	TELWH	Early Write Pulse Hold Time	80		40	50	ns	④
	TDVWL	Data Setup Time	0		-15	-10	ns	④
	TDVEL	Early Write Data Setup Time	0		-15	-10	ns	④
	TWLDX	Data Hold Time	80		50	60	ns	④
	TELDX	Early Write Data Hold Time	80		50	60	ns	④
	TELWL	Early Write Output Hi-Z Time	0		-10	-10	ns	④
TQVWL	Data Valid to Write Time	0		0	0	ns	④	
TELEL	Read or Write Cycle Time	420		190	260	ns	④	

- NOTES:
- All devices guaranteed at worst case limits. Room temp., 5 volt data provided for information and not guaranteed.
  - Operating Supply Current (ICCOP) is proportional to Operating Frequency. Post Rad Data @ TD = 1 x 10<sup>5</sup>.
  - Capacitance sampled and guaranteed - not 100% tested.
  - AC test Conditions: Inputs: TRISE = TFALL < 20nsec; Outputs: 1TTL Load and 50 pF. All timing measurements at 1/2 VCC.
  - Pre-Radiation and Post-Radiation limits.

### Read Cycle



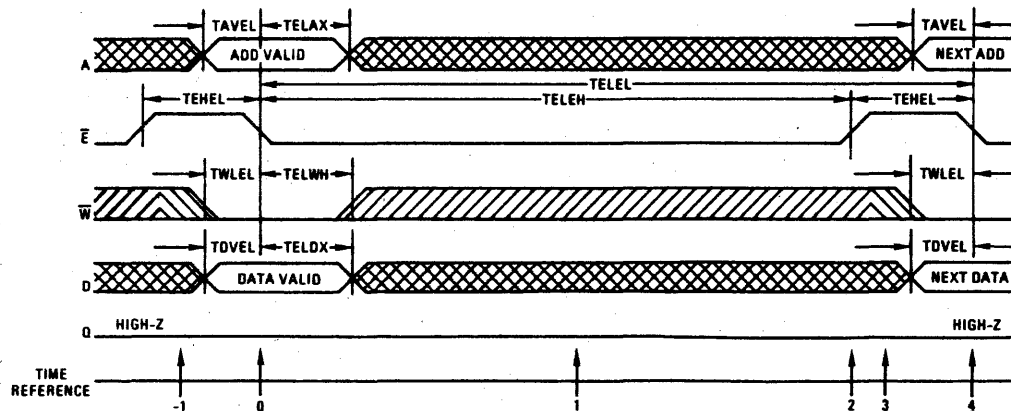
TRUTH TABLE

TIME REFERENCE	INPUTS			OUTPUT	FUNCTION
	$\bar{E}$	$\bar{W}$	A	Q	
-1	H	X	X	Z	MEMORY DISABLED
0	L	H	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	H	X	X	OUTPUT ENABLED
2	L	H	X	V	OUTPUT VALID
3	H	X	X	V	READ ACCOMPLISHED
4	H	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	L	H	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The address information is latched in the on chip registers on the falling edge of  $\bar{E}$  ( $T = 0$ ). Minimum address set up and hold time requirements must be met. After the required hold time, the addresses may change state without affecting device operation. During time ( $T = 1$ ) the output

becomes enabled but data is not valid until during time ( $T = 2$ ).  $\bar{W}$  must remain high until after time ( $T = 2$ ). After the output data has been read,  $\bar{E}$  may return high ( $T = 3$ ). This will disable the output buffer and ready the RAM for the next memory cycle ( $T = 4$ ).

### Early Write Cycle



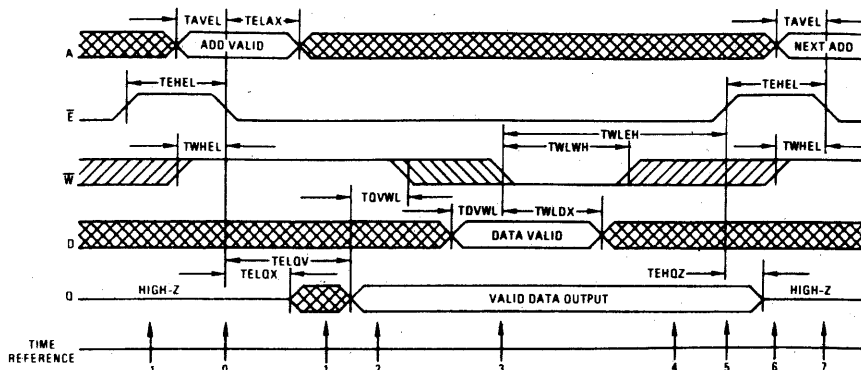
TRUTH TABLE

TIME REFERENCE	INPUTS				OUTPUT	FUNCTION
	$\bar{E}$	$\bar{W}$	A	D	Q	
-1	H	X	X	X	Z	MEMORY DISABLED
0	L	L	V	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	X	X	X	Z	WRITE IN PROGRESS INTERNALLY
2	H	X	X	X	Z	WRITE COMPLETED
3	H	X	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
4	L	L	V	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The early write cycle is the only cycle where the output is guaranteed not to become active. On the falling edge of  $\bar{E}$  ( $T = 0$ ), the addresses; the write signal, and the data input are latched in on chip registers. The logic value of  $\bar{W}$  at the time  $\bar{E}$  falls determines the state of the output buffer for that cycle. Since  $\bar{W}$  is low when  $\bar{E}$  falls, the output buffer is latched into the high impedance state and

will remain in that state until  $\bar{E}$  returns high ( $T = 2$ ). For this cycle, the data input is latched by  $\bar{E}$  going low; therefore data set up and hold times should be referenced to  $\bar{E}$ . When  $\bar{E}$  ( $T = 2$ ) returns to the high state the output buffer disables and all signals are unlatched. The device is now ready for the next cycle.

### Read Modify Write Cycle



TRUTH TABLE

TIME REFERENCE	E	W	A	D	OUTPUT O	FUNCTION
-1	H	X	X	X	Z	MEMORY DISABLED
0	↘	H	V	X	Z	CYCLE BEGINS, ADDRESS ARE LATCHED
1	L	H	X	X	X	OUTPUT ENABLED
2	L	H	X	X	V	OUTPUT VALID, READ AND MODIFY TIME
3	L	↘	X	V	V	WRITE BEGINS, DATA IS LATCHED
4	L	X	X	X	V	WRITE IN PROGRESS INTERNALLY
5	↘	X	X	X	V	WRITE COMPLETED
6	H	X	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
7	↘	H	V	X	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The read modify write cycle begins as all other cycles on the falling edge of  $\bar{E}$  (T = 0). The  $\bar{W}$  line should be high at (T = 0) in order to latch the output buffers in the active state. During (T = 1) the output will be active but not valid until (T = 2). On the falling edge of the  $\bar{W}$  (T = 3) the data present at the output and input are latched. The

$\bar{W}$  signal also latches itself on its low going edge. All input signals excluding  $\bar{E}$  have been latched and have no further effect on the RAM. The rising edge of  $\bar{E}$  (T = 5) completes the write portion of the cycle and unlatches all inputs and output. The output goes to a high impedance and the RAM is ready for the next cycle.

NOTES: In the above descriptions the numbers in parenthesis (T = n) refer to the respective timing diagrams. The numbers are located on the time reference line below each diagram. The timing diagrams shown are only examples and are not the only valid method of operation.

### Radiation Screening Procedure

- Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- The sample die shall be assembled and tested to the production test program for proper operation.
- The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^5$  Rad Si ( $\pm 10\%$ ) from a Gamma Cell 220 Cobalt 60 source or equivalent. The samples shall be biased at 5 volts with all inputs high. The dose rate shall be between 50 rads/sec and 200 rads/sec.
- The samples will be tested to the data sheet limits within one hour ( $\pm 15$  minutes) after irradiation. The lot will be accepted only if all units, exclusive of non-radiation failures, meet the data sheet limits.

### Radiation Effects

The HS-6504RH is an RH memory designed to survive in a radiation environment and to meet the electrical characteristics and be pin compatible to the Harris equivalent commercial part. Latchup free operation, achieved by the use of special starting material and improved total dose hardness, is obtained with special high temperature processing cycles. These process techniques can, in principle, be applied to any standard HARRIS CMOS product.

On a production basis, HARRIS only performs screens for total dose hardness to a level of  $1 \times 10^5$  rad-Si. Transient radiation tests, however, have shown the following results:

- Latchup free to doses  $\geq 5 \times 10^{11}$  rads/sec.
- Upset (loss of stored data) typically  $\geq 10^8$  rads/sec.

**Test Product Flow**

**HARRIS SEMICONDUCTOR PRODUCT FLOW  
MIL-STD-883, METHOD 5004 CLASS B**

**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

**Notes:** Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.

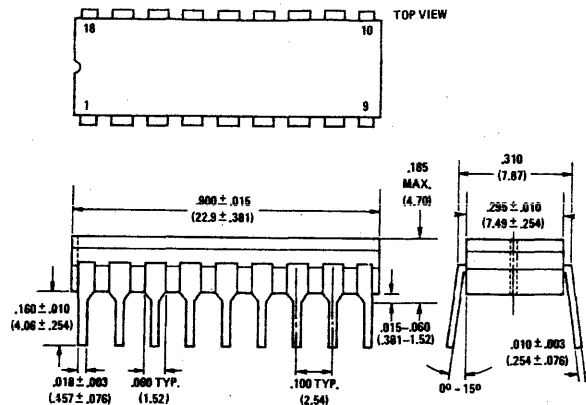
Branding: All devices are branded with the part number and EIA date code.

Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.

Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.

**Packaging**

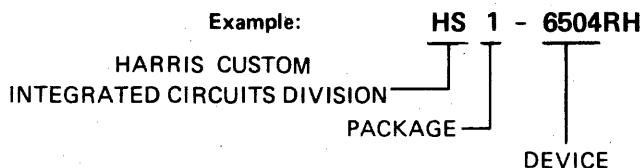
**18 LEAD CERAMIC DIP**



1. All dimensions in inches; millimeters are shown in parentheses.
2. All dimensions  $\pm .010$  ( $\pm 0.25$ mm) unless otherwise shown.



## Ordering Information



### PACKAGE

FLAT PACK	9-
CERDIP	1-

*NOTICE: Harris Semiconductor's products are sold by description only. Harris Semiconductor reserves the right to make changes in circuit design and/or specifications at any time without notice. Accordingly, the reader is cautioned to verify that data sheets are current before placing orders.*

### Sales Offices

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##### EASTERN REGION

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
TWX: 710-332-1074

106 Seventh Street  
Garden City, NY 11530  
(516) 747-6776

##### WESTERN REGION

1503 South Coast Drive  
Suite 320  
Costa Mesa, CA 92626  
(714) 957-6557  
TWX: 910-595-1533

625 Ellis, Suite 300  
Mt. View, CA 94043  
(415) 964-6443  
TWX: 910-379-6431

##### MIDWEST REGION

2850 Metro Dr., Suite 703  
Minneapolis, MN 55420  
(612) 854-3224  
TWX: 910-576-3418

##### HOME OFFICE

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

#### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



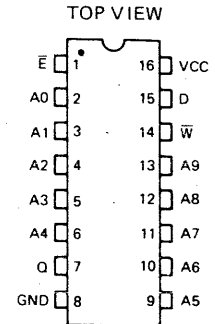
# HARRIS

CUSTOM INTEGRATED CIRCUITS DIVISION

### Features

- FUNCTIONAL TOTAL DOSE .....  $2 \times 10^4$  RAD Si
- LATCH-UP FREE TO .....  $5.0 \times 10^{11}$  RAD Si/sec
- LOW STANDBY POWER ..... 550  $\mu$ W MAX
- LOW OPERATING POWER ..... 25mW/MHz MAX
- FAST ACCESS TIME ..... 300nsec MAX
- TTL COMPATIBLE IN/OUT
- HIGH OUTPUT DRIVE – 2 TTL LOADS
- HIGH NOISE IMMUNITY
- ON-CHIP ADDRESS REGISTER
- MILITARY TEMPERATURE RANGE
- THREE-STATE OUTPUTS
- 16 PIN PACKAGE FOR HIGH DENSITY

### Pinout



A – Address Input    D – Data Input  
 $\bar{E}$  – Chip Enable    Q – Data Output  
 W – Write Enable

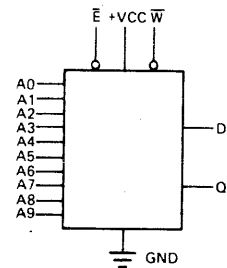
### Description

The HS-6508RH is a 1024 by 1 static CMOS RAM fabricated using the HARRIS Programs Division radiation hardened self-aligned silicon gate technology. Synchronous circuit design techniques are employed to achieve high performance and low power operation.

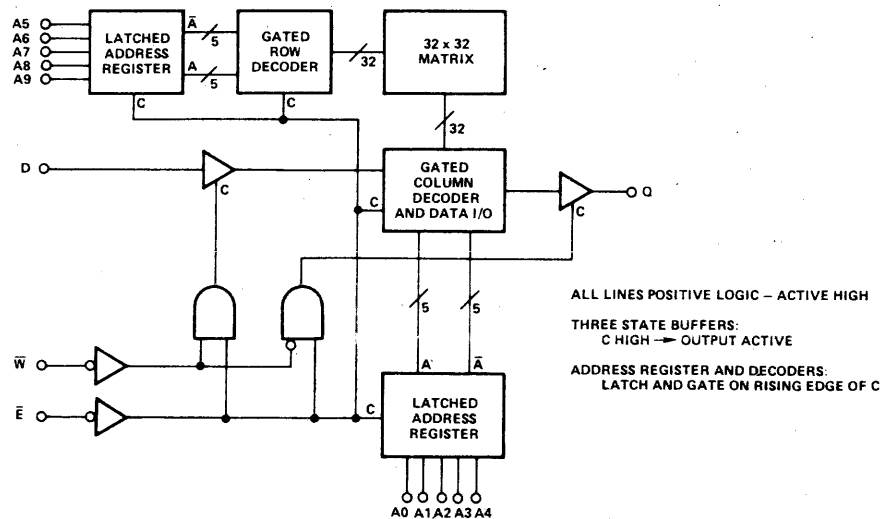
On-chip latches are provided for addresses allowing efficient interfacing with microprocessor systems. The data output buffers can be forced to a high impedance state for use in expanded memory arrays.

The HS-6508RH is a fully static RAM and may be maintained in any state for an indefinite period of time.

### Logic Symbol



### Functional Diagram



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## Symbols and Abbreviations

This data sheet utilizes a new set of specification nomenclature. This new format is an IEEE and JEDEC supported standard for semiconductor memories. It is intended to clarify the symbols, abbreviations and definitions, and to make all memory data sheets consistent. We believe that, once acclimated, you will find this standardized format easy to read and use.

### ELECTRICAL PARAMETER ABBREVIATIONS

All abbreviations use upper case letters with no subscripts. The initial symbol is one of these four characters:

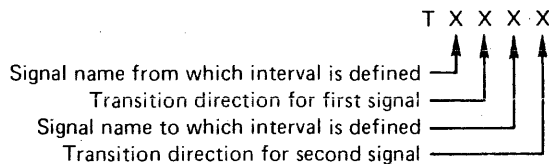
- V (Voltage)
- I (Current)
- P (Power)
- C (Capacitance)

The second letter specifies input (I) or output (O), and the third letter indicates the high (H), low (L) or off (Z) state of the pin during measurements. Examples:

- VIL — Input Low Voltage
- IOZ — Output Leakage Current

### TIMING PARAMETER ABBREVIATIONS

All timing abbreviations use upper case characters with no subscripts. The initial character is always T and is followed by four descriptors. These characters specify two signal points arranged in a "from-to" sequence that define a timing interval. The two descriptors for each signal point specify the signal name and the signal transitions. Thus the format is:



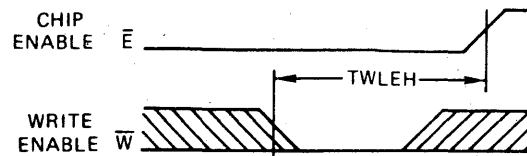
Signal Definitions:

- A = Address
- D = Data In
- Q = Data Out
- W = Write Enable
- E = Chip Enable
- S = Chip Select
- G = Output Enable

Transition Definitions:

- H = Transition to High
- L = Transition to Low
- V = Transition to Valid
- X = Transition to Invalid or Don't Care
- Z = Transition to Off (High Impedance)

EXAMPLE:



The example shows Write pulse setup time defined as TWLEH—Time from Write enable Low to chip Enable High.

### TIMING LIMITS

The table of timing values shows either a minimum or a maximum limit for each parameter. Input requirements are specified from the external system point of view. Thus, address set-up time is shown as a minimum since the system must supply at least that much time (even though most devices do not require it). On the other hand, responses from the memory are specified from the device point of view. Thus, the access time is shown as a maximum since the device never provides data later than that time.

### WAVEFORMS

WAVEFORM SYMBOL	INPUT	OUTPUT
	MUST BE VALID	WILL BE VALID
	CHANGE FROM H TO L	WILL CHANGE FROM H TO L
	CHANGE FROM L TO H	WILL CHANGE FROM L TO H
	DON'T CARE: ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
	—	HIGH IMPEDANCE



ABSOLUTE MAXIMUM RATINGS		OPERATING RANGE	
Supply Voltage -VCC	7.0V	Operating Supply Voltage -VCC	4.5V to 5.5V
Input or Output Voltage Applied	GND -0.3V to VCC +0.3V	Operating Temperature	-55°C to +125°C
Storage Temperature	-65°C to +150°C		

**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

ELECTRICAL CHARACTERISTICS <sup>⑤</sup>

D.C.

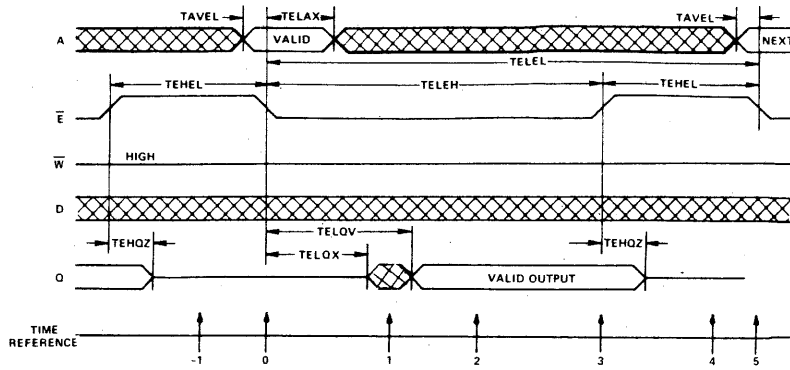
SYMBOL	PARAMETER	TEMP. & VCC = OPERATING RANGE		TEMP. = 25°C <sup>①</sup> VCC = 5.0V			UNITS	TEST CONDITIONS
		MIN	MAX	MIN	TYP	MAX		
ICCSB	Standby Supply Current		100		10	100	μA	IO = 0 VI = VCC or GND
ICCOP	Operating Supply Current <sup>②</sup>		4		1.5	2.5	mA	f = 1MHz, IO = 0 VI = VCC or GND
II	Input Leakage Current	-1.0	+1.0	-0.5	0.0	+0.5	μA	GND ≤ VI ≤ VCC
IOZ	Output Leakage Current	-1.0	+1.0	-0.5	0.0	+0.5	μA	GND ≤ VI ≤ VCC
VIL	Input Low Voltage	-0.3	0.8	-0.3		1.5	V	
VIH	Input High Voltage	VCC -2.0	VCC +0.3	2.5		5.3	V	
VOL	Output Low Voltage		0.4		0.2	0.35	V	IOL = 3.2mA
VOH	Output High Voltage	2.4		3.0	4.5		V	IOH = -3mA
CI	Input Capacitance <sup>③</sup>		6		4	6	pF	VI = VCC or GND f = 1MHz
CO	Output Capacitance <sup>③</sup>		10		6	10	pF	VO = VCC or GND f = 1MHz

A.C.

TELQV	Chip Enable Access Time		300		160	250	ns	④
TAVQV	Address Access Time		310		160	260	ns	④
TELOX	Chip Enable Output Enable Time		200		60	170	ns	④
TWLOZ	Write Enable Output Disable Time		200		60	170	ns	④
TEHQZ	Chip Enable Output Disable Time		200		60	170	ns	④
TELEH	Chip Enable Pulse Negative Width	300		250	160		ns	④
TEHEL	Chip Enable Pulse Positive Width	150		130	90		ns	④
TAVEL	Address Setup Time	10		10	0		ns	④
TELAX	Address Hold Time	70		50	40		ns	④
TDVWH	Data Setup Time	130		100	80		ns	④
TWHDX	Data Hold Time	0		0	0		ns	④
TWLEH	Chip Enable Write Pulse Setup Time	160		130	100		ns	④
TELWH	Chip Enable Write Pulse Hold Time	160		130	100		ns	④
TWLWH	Write Enable Pulse Width	160		130	100		ns	④
TELEL	Read or Write Cycle Time	450		380	250		ns	④

- NOTES:
- All devices guaranteed at worst case limits. Room temp., 5 volt data provided for information and not guaranteed.
  - Operating Supply Current (ICCOP) is proportional to Operating Frequency. Example: Typical ICCOP = 1.5mA/MHz.
  - Capacitance sampled and guaranteed – not 100% tested.
  - AC Test Conditions: Inputs – TRISE = TFALL = 20nsec; Outputs – 1 TTL load and 50pF. All timing measurements at 1/2 VCC.
  - Pre-Radiation characteristics. See Radiation effects section for Post-Radiation characteristics.

### Read Cycle



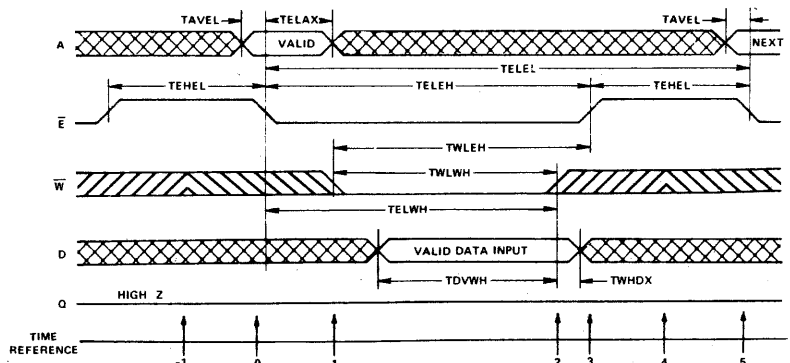
TRUTH TABLE

TIME REFERENCE	$\bar{E}$	INPUTS W A D	OUTPUTS Q	FUNCTION
-1	H	X X X	Z	MEMORY DISABLED
0	L	H V X	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	H X X	X	OUTPUT ENABLED
2	L	H X X	V	OUTPUT VALID
3	H	H X X	V	READ ACCOMPLISHED
4	H	X X X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	H	H V X	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

In the HS-6508RH Read Cycle, the address information is latched into the on-chip registers on the falling edge of  $\bar{E}$  (T = 0). Minimum address setup and hold time requirements must be met. After the required hold time, the addresses may change state without affecting device operation. During time (T = 1) the data output becomes enabled; however, the data is not valid until during time

(T = 2).  $\bar{W}$  must remain high for the read cycle. After the output data has been read,  $\bar{E}$  may return high (T = 3). This will disable the chip and force the output buffer to a high impedance state. After the required  $\bar{E}$  high time (TEHEL) the RAM is ready for the next memory cycle (T = 4).

### Write Cycle



TRUTH TABLE

TIME REFERENCE	$\bar{E}$	INPUTS W A D	OUTPUTS Q	FUNCTION
-1	H	X X X	Z	MEMORY DISABLED
0	L	X V X	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	X X X	Z	WRITE PERIOD BEGINS
2	L	X V X	Z	DATA IS WRITTEN
3	H	X X X	Z	WRITE COMPLETED
4	H	X X X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	H	X V X	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)



The write cycle is initiated by the falling edge of  $\bar{E}$  which latches the address information into the on-chip registers. The write portion of the cycle is defined as both  $\bar{E}$  and  $\bar{W}$  being low simultaneously.  $\bar{W}$  may go low anytime during the cycle provided that the write enable pulse setup time (TWLEH) is met. The write portion of the cycle is terminated by the first rising edge of either  $\bar{E}$  or  $\bar{W}$ . Data setup and hold times must be referenced to the terminating signal.

If a series of consecutive write cycles are to be performed, the  $\bar{W}$  line may remain low until all desired locations have been written. When this method is used, data setup and hold times must be referenced to the rising edge of  $\bar{E}$ . By

positioning the  $\bar{W}$  pulse at different times within the  $\bar{E}$  low time (TELEH), various types of write cycles may be performed.

If the  $\bar{E}$  low time (TELEH) is greater than the  $\bar{W}$  pulse (TWLWH) plus an output enable time (TELQX), a combination read write cycle is executed. Data may be modified an indefinite number of times during any write cycle (TELEH). The data input and data output pins may be tied together for use with a common I/O data bus structure. When using the RAM in this method allow a minimum of one output disable time (TWLQZ) after  $\bar{W}$  goes low before applying input data to the bus. This will insure that the output buffers are not active.

### Radiation Screening Procedure

(1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)

(2) The sample die shall be assembled and tested for functionality.

(3) The sample devices shall be subjected to a Total

Dose Radiation level of  $2 \times 10^4$  Rad Si ( $\pm 10\%$ ) from a Gamma Cell 220 Cobalt 60 source or equivalent. The samples shall be biased at 5 volts with all inputs high. The dose rate shall be between 50 rads/sec and 200 rads/sec.

(4) ICCSB at  $V_{CC} = 5$  volts will be measured and recorded for each device within one hour ( $\pm 15$  minutes) after irradiation. The lot will be accepted only if the average of these measured values is  $\leq 10$  mA.

### Radiation Effects

The HS-6508RH is a radiation hardened memory processed with the same mask set as is used for HARRIS' equivalent commercial part. Latchup free operation, achieved by the use of special starting material and improved total dose hardness, is obtained with special high temperature processing cycles. These process techniques can, in principal, be applied to any standard HARRIS CMOS product.

The primary failure mode under exposure to ionizing radiation is an increase in static leakage current (ICCSB). Functional failure due to the increased leakage currents will typically occur for dose levels

in excess of  $5 \times 10^4$  RAD-SI. AC and DC parameters other than ICC will change less than 10% for total dose levels under  $5 \times 10^4$  RAD-SI. The excess leakage currents will anneal at room temperature and are typically reduced by a factor of 3-10 within 24 hours after irradiation.

On a production basis, HARRIS is able to perform screens only for total dose hardness. Transient radiation tests, however, have shown the following results:

Latchup free to doses  $\geq 5 \times 10^{11}$  rads/sec.

Upset (loss of stored data) typically  $\geq 10^8$  rads/sec.

**HARRIS SEMICONDUCTOR PRODUCT FLOW**  
**MIL-M-38510/MIL-STD-883, METHOD 5004 CLASS B**

**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: (A) Fine (B) Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

Note:

Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.

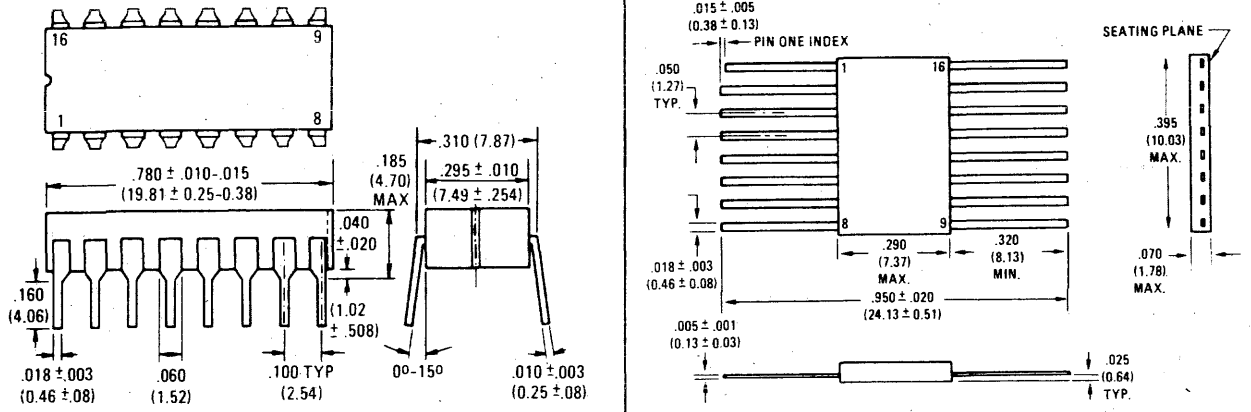
Branding: All devices are branded with the part number and EIA date code.

Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.

Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.

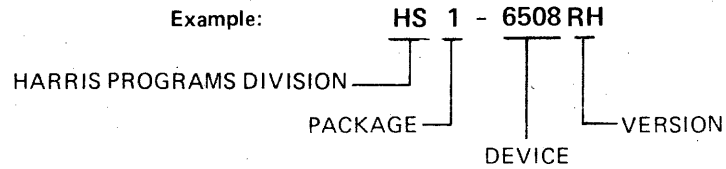


### Packaging



1. All dimensions in inches; millimeters are shown in parentheses.
2. All dimensions  $\pm .010$  ( $\pm 0.25$ mm) unless otherwise shown.

### Ordering Information



#### PACKAGE

FLAT PACK	9-
CERDIP	1-

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# HARRIS

## HS-6514RH

### 1024 x 4 CMOS RAM

*Preliminary*

#### Features

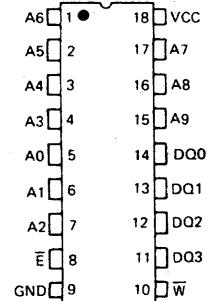
- LOW POWER STANDBY
- LOW POWER OPERATION
- FUNCTIONAL TOTAL DOSE
- DATA UPSET
- LATCH - UP FREE TO
- TTL COMPATIBLE INPUT/OUTPUT
- COMMON DATA IN/OUT
- THREE - STATE OUTPUTS
- STANDARD JEDEC PINOUT
- FAST ACCESS TIME
- MILITARY TEMPERATURE RANGE
- 18 PIN PACKAGE FOR HIGH DENSITY
- ON - CHIP ADDRESS REGISTER

100  $\mu$ W MAX.  
 25mW/MHz MAX.  
 $1 \times 10^5$  RAD Si  
 $>10^8$  RADS Si/sec  
 $>5 \times 10^{11}$  RAD Si/sec

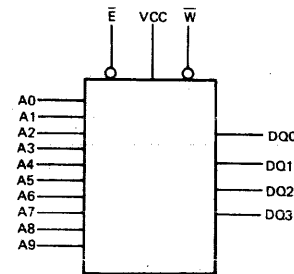
200 nsec TYP.

#### Pinout

TOP VIEW



#### Logic Symbol



A - Address Input  
 E - Chip Enable  
 W - Write Enable  
 DQ - Data In/Out

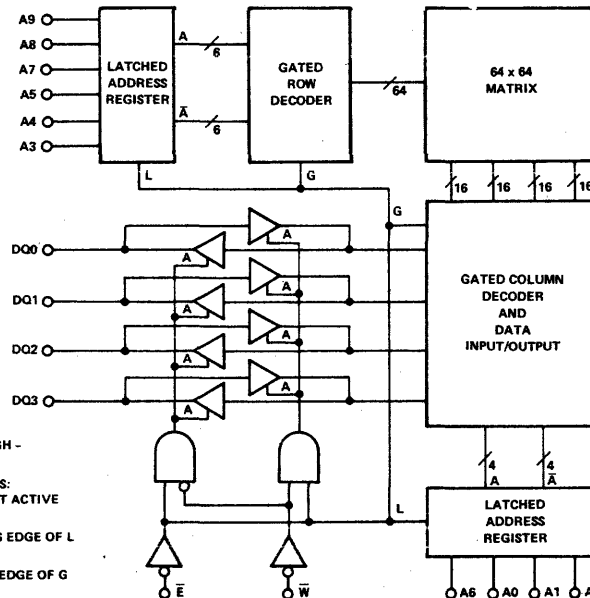
#### Description

The HS-6514RH is a 1024 x 4 static CMOS RAM fabricated using the Harris Custom Integrated Circuits Division radiation hardened self-aligned silicon gate technology. The device utilizes synchronous circuitry to achieve high performance and low power operation.

On-chip latches are provided for the addresses allowing efficient interfacing with microprocessor systems. The data output can be forced to a high impedance state for use in expanded memory systems.

The HS-6514RH is a fully static RAM and may be maintained in any state for an indefinite period of time.

#### Functional Diagram



ALL LINES ACTIVE HIGH - POSITIVE LOGIC  
 THREE STATE BUFFERS: A HIGH  $\rightarrow$  OUTPUT ACTIVE  
 ADDRESS REGISTERS: LATCH ON RISING EDGE OF L  
 GATED DECODERS: GATE ON RISING EDGE OF G

Information on this device is preliminary. Data is subject to change unless otherwise specifically agreed. No obligations are assumed for notice of change or future manufacture of this device.

CAUTION: These devices are sensitive to electrostatic discharge.

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Harris Semiconductor

CUSTOM/SEMICUSTOM



## Symbols and Abbreviations

This data sheet utilizes a new set of specification nomenclature. This new format is an IEEE and JEDEC supported standard for semiconductor memories. It is intended to clarify the symbols, abbreviations and definitions, and to make all memory data sheets consistent. We believe that, once acclimated, you will find this standardized format easy to read and use.

### ELECTRICAL PARAMETER ABBREVIATIONS

All abbreviations use upper case letters with no subscripts. The initial symbol is one of these four characters:

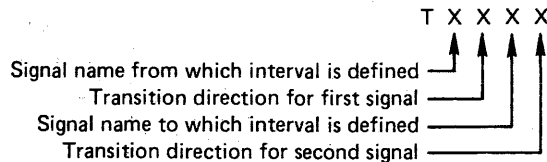
- V (Voltage)
- I (Current)
- P (Power)
- C (Capacitance)

The second letter specifies input (I) or output (O), and the third letter indicates the high (H), low (L) or off (Z) state of the pin during measurements. Examples:

- VIL — Input Low Voltage
- IOZ — Output Leakage Current

### TIMING PARAMETER ABBREVIATIONS

All timing abbreviations use upper case characters with no subscripts. The initial character is always T and is followed by four descriptors. These characters specify two signal points arranged in a "from-to" sequence that define a timing interval. The two descriptors for each signal point specify the signal name and the signal transitions. Thus the format is:



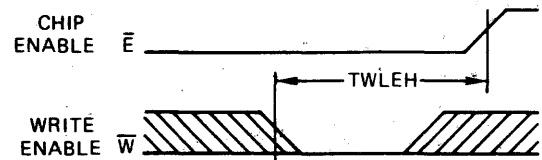
#### Signal Definitions:

- A = Address
- D = Data In
- O = Data Out
- W = Write Enable
- E = Chip Enable
- S = Chip Select
- G = Output Enable

#### Transition Definitions:

- H = Transition to High
- L = Transition to Low
- V = Transition to Valid
- X = Transition to Invalid or Don't Care
- Z = Transition to Off (High Impedance)

### EXAMPLE:



The example shows Write pulse setup time defined as TWLEH—Time from Write enable Low to chip Enable High.

### TIMING LIMITS

The table of timing values shows either a minimum or a maximum limit for each parameter. Input requirements are specified from the external system point of view. Thus, address set-up time is shown as a minimum since the system must supply at least that much time (even though most devices do not require it). On the other hand, responses from the memory are specified from the device point of view. Thus, the access time is shown as a maximum since the device never provides data later than that time.

### WAVEFORMS

WAVEFORM SYMBOL	INPUT	OUTPUT
	MUST BE VALID	WILL BE VALID
	CHANGE FROM H TO L	WILL CHANGE FROM H TO L
	CHANGE FROM L TO H	WILL CHANGE FROM L TO H
	DON'T CARE: ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
	—	HIGH IMPEDANCE

# Specifications HS-6514RH

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage — (VCC - GND)	-0.3V to +7.0V
Input or Output Voltage Applied	(GND -0.3V) to (VCC + 0.3V)
Storage Temperature	-65°C to +150°C

## OPERATING RANGE

Operating Supply Voltage	4.5V to 5.5V
Operating Temperature	-55°C to +125°C

**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

## ELECTRICAL CHARACTERISTICS ⑤

D.C.

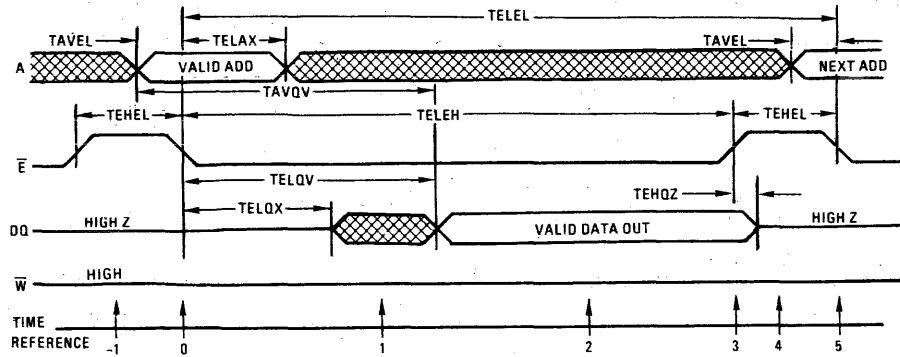
SYMBOL	PARAMETER	TEMP. & VCC = OPERATING RANGE		TEMP = 25°C ① VCC = 5.0V TYPICAL		UNITS	TEST CONDITIONS
		MIN	MAX	PRE RAD	POST RAD		
ICCSB	Standby Supply Current		250	6	6	μA	IO = 0 VI = VCC or GND
ICCOP	Operating Supply Current ②		7	4.5		mA	t = 1MHz, IO = 0 VI = VCC or GND
ICCDR	Data Retention Current		50	4	4	μA	IO=0 VI=VCC or GND VCC=3.0V
VCCDR	Data Retention Voltage		3.0	2.6	1.8	V	
II	Input Leakage Current	-1	+1	0	0	μA	GND ≤ VI ≤ VCC
IIOZ	Input/Output Leakage Current	-10	+10	+0.5	+0.5	μA	GND ≤ VIO ≤ VCC
VIL	Input Low Voltage	-0.3	0.8	1.5	1.3	V	
VIH	Input High Voltage	VCC - 2.0	VCC + 0.3	2.7	2.3	V	
VOL	Output Low Voltage		0.40	0.2	0.15	V	IO = 2.0mA
VOH	Output High Voltage	2.4		4.7	4.1	V	IO = -1.0mA
CI	Input Capacitance ③		8.0	5.0	5.0	pF	VI = VCC or GND f = 1MHz
CIO	Input/Output Capacitance ③		10.0	6.0	6.0	pF	VIO = VCC or GND f = 1MHz

A.C.

TELOV	Chip Enable Access Time		300	170	200	ns	④
TAVQV	Address Access Time		320	170	200	ns	④
TELOX	Chip Enable Output Enable Time		100	40	50	ns	④
TWLOZ	Write Enable Output Disable Time	20	100	40	50	ns	④
TEHQZ	Chip Enable Output Disable Time		100	40	50	ns	④
TELEH	Chip Enable Pulse Negative Width	300		170	200	ns	④
TEHEL	Chip Enable Pulse Positive Width	120		40	80	ns	④
TAVEL	Address Setup Time	20		-10	0	ns	④
TELAX	Address Hold Time	100		50	60	ns	④
TWLWH	Write Enable Pulse Width	300		150	150	ns	④
TWLEH	Write Enable Pulse Setup Time	300		150	150	ns	④
TELWH	Write Enable Pulse Hold Time	300		150	150	ns	④
TDVWH	Data Setup Time	200		100	120	ns	④
TWHDV	Data Hold Time	30		10	15	ns	④
TWLDV	Write Data Delay Time	100		50	60	ns	④
TWLEL	Early Output High-Z Time	0		-20	-15	ns	④
TEHWH	Late Output High-Z Time	0		-20	-15	ns	④
TELEL	Read or Write Cycle Time	420		210	280	ns	④

- NOTES: 1. All devices guaranteed at worst case limits. Room temp., 5 volt data provided for information and not guaranteed.  
 2. Operating Supply Current (ICCOP) is proportional to Operating Frequency. Post Rad data at TD = 1 X 10<sup>5</sup>.  
 3. Capacitance sampled and guaranteed — not 100% tested.  
 4. AC test Conditions: Inputs: TRISE = TFALL ≤ 20nsec; Outputs: 1 TTL Load and 50pF. All timing measurements at 1/2 VCC.  
 5. Pre-Radiation and Post-Radiation limits.

### Read Cycle



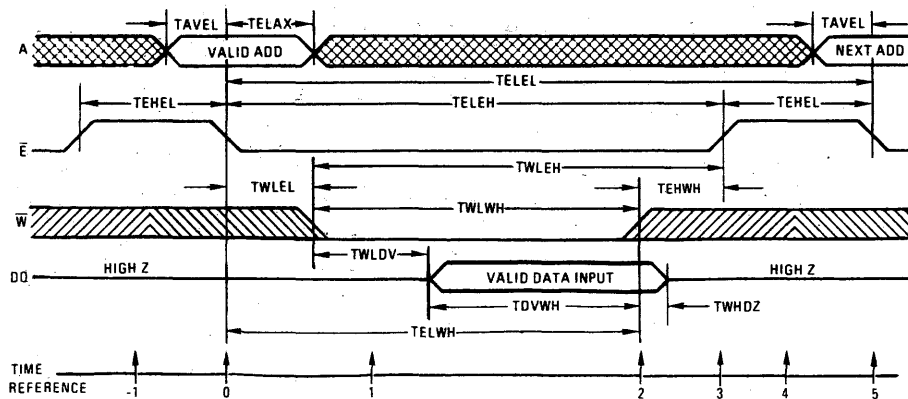
TRUTH TABLE

TIME REFERENCE	INPUTS			DATA I/O	FUNCTION
	$\bar{E}$	$\bar{W}$	A	DQ	
-1	H	X	X	Z	MEMORY DISABLED
0	L	H	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	H	X	X	OUTPUT ENABLED
2	L	H	X	V	OUTPUT VALID
3	L	H	X	V	READ ACCOMPLISHED
4	H	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	L	H	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The address information is latched in the on chip registers on the falling edge of  $\bar{E}$  ( $T = 0$ ). Minimum address setup and hold time requirements must be met. After the required hold time the addresses may change state without affecting device operation. During time ( $T = 1$ ) the outputs become enabled but data is not valid until time ( $T = 2$ ).

$\bar{W}$  must remain high throughout the read cycle. After the data has been read  $\bar{E}$  may return high ( $T = 3$ ). This will force the output buffers into a high impedance mode at time ( $T = 4$ ). The memory is now ready for the next cycle.

### Write Cycle



TRUTH TABLE

TIME REFERENCE	INPUTS			DATA I/O	FUNCTION
	$\bar{E}$	$\bar{W}$	A	DQ	
-1	H	X	X	Z	MEMORY DISABLED
0	L	X	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	L	X	Z	WRITE PERIOD BEGINS
2	L	L	X	V	DATA IN IS WRITTEN
3	L	H	X	Z	WRITE COMPLETED
4	H	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	L	X	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The write cycle is initiated by the falling edge of  $\bar{E}$  ( $T = 0$ ), which latches the address information in the on chip registers.

There are two basic types of write cycles, which differ in the control of the common data-in/data-out bus.

**Case 1:  $\bar{E}$  falls before  $\bar{W}$  falls**

The output buffers may become enabled (reading) if  $\bar{E}$  falls before  $\bar{W}$  falls.  $\bar{W}$  is used to disable (three-state) the outputs so input data can be applied. TWLDV must be met to allow the  $\bar{W}$  signal time to disable the outputs before applying input data. Also, at the end of the cycle the outputs may become active if  $\bar{W}$  rises before E. The RAM outputs will disable (three-state) after E rises (TEHQZ). In this type of write cycle TWLEL and TEHWH may be ignored.

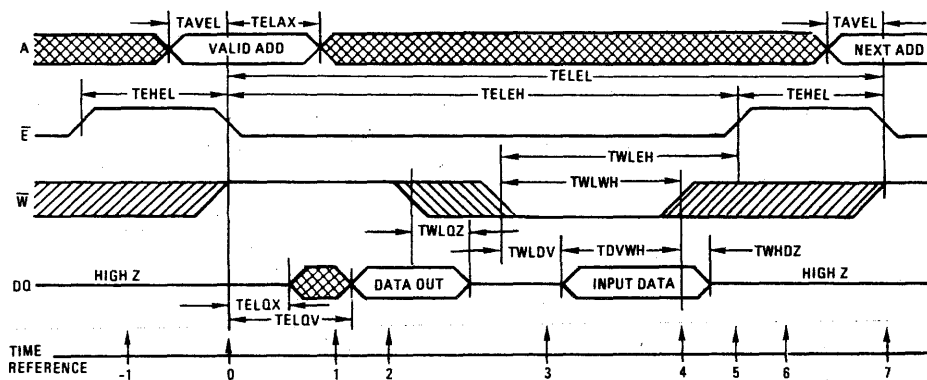
**Case 2:  $\bar{E}$  falls equal to or after  $\bar{W}$  falls, and  $\bar{E}$  rises before or equal to  $\bar{W}$  rises.**

This  $\bar{E}$  and  $\bar{W}$  control timing will guarantee that the data outputs will stay disabled throughout the cycle, thus simplifying the data input timing. TWLEL and TEHWH must be met but TWLDV becomes meaningless and can be ignored. In this cycle TDVWH and TWHZD become TDVEH and TEHDZ. In other words, reference data setup and hold times to the  $\bar{E}$  rising edge.

If a series of consecutive write cycles are to be performed,  $\bar{W}$  may be held low until all desired locations have been written (an extension of Case 2).

	IF	OBSERVE	IGNORE
Case 1	$\bar{E}$ falls before $\bar{W}$	TWLDV	TWLEL
Case 2	$\bar{E}$ falls after $\bar{W}$ & $\bar{E}$ rises before $\bar{W}$	TWLEL TEHWH	TWLDV TWHZD

**Read Modify Write Cycle**



**TRUTH TABLE**

TIME REFERENCE	INPUTS			DATA/I/O	FUNCTION
	$\bar{E}$	$\bar{W}$	A	DO	
-1	H	X	X	Z	MEMORY DISABLED
0	L	H	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	H	X	X	READ MODE, OUTPUT ENABLED
2	L	H	X	V	READ MODE, OUTPUT VALID
3	L	L	X	Z	WRITE MODE, OUTPUT HIGH Z
4	L	L	X	V	WRITE MODE, DATA IS WRITTEN
5	L	H	X	Z	WRITE COMPLETED
6	H	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
7	H	H	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

If the pulse width of  $\bar{W}$  is relatively short in relation to that of  $\bar{E}$  a combination read-write cycle may be performed. If  $\bar{W}$  remains high for the first part of the cycle, the outputs will become active during time (T = 1). Data out will be valid during time (T = 2). After the data is read,  $\bar{W}$  can go low. After minimum TWLWH,  $\bar{W}$  may return high. The

information just written may now be read or  $\bar{E}$  may return high, disabling the output buffers and preparing the device for the next cycle. Any number or sequence of read-write operations may be performed while  $\bar{E}$  is low providing all timing requirements are met.

**NOTES:**

In the above descriptions the numbers in parenthesis (T = n) refer to the respective timing diagrams. The numbers are located on the time reference line below each diagram. The timing diagrams shown are only examples and are not the only valid method of operation.



### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested to the production test program for proper operation.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^5$  Rad Si ( $\pm 10\%$ ) from a Gamma Cell 220 Cobalt 60 source or equivalent. The samples shall be biased at 5 volts with all inputs high. The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4) The samples will be tested to the data sheet limits within one hour ( $\pm 15$  minutes) after irradiation. The lot will be accepted only if all units, exclusive of non-radiation failures, meet the data sheet limits.

### Radiation Effects

The HS-6514RH is an RH memory designed to survive in a radiation environment and to meet the electrical characteristics and be pin compatible to the Harris equivalent commercial part. Latchup free operation, achieved by the use of special starting material and improved total dose hardness, is obtained with special high temperature processing cycles. These process techniques can, in principle, be applied to any standard HARRIS CMOS product.

On a production basis, HARRIS only performs screens for total dose hardness to a level of  $1 \times 10^5$  rad-Si. Transient radiation tests, however, have shown the following results:

- Latchup free to doses  $\geq 5 \times 10^{11}$  rads/sec.
- Upset (loss of stored data) typically  $\geq 10^8$  rads/sec.

### Test Product Flow

#### HARRIS SEMICONDUCTOR PRODUCT FLOW MIL-STD-883, METHOD 5004 CLASS B

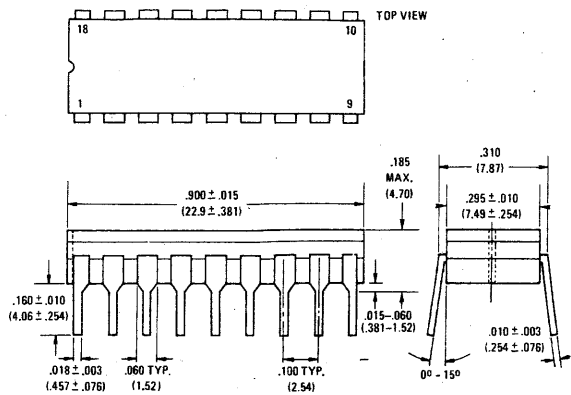
#### 100% SCREENING PROCEDURE

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

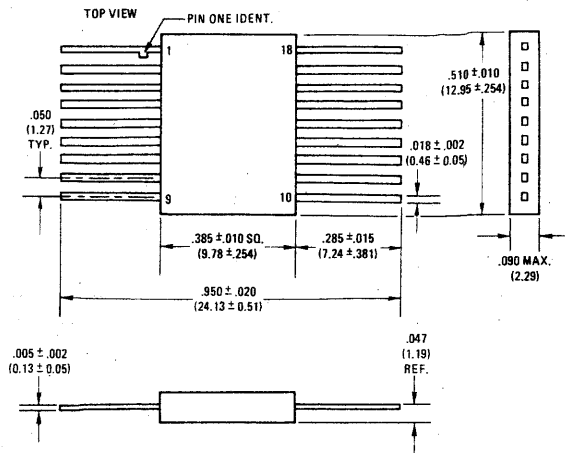
- Notes:**
- Traceability:** All devices are assigned date code identification that provides traceability back to the inspection lot.
  - Branding:** All devices are branded with the part number and EIA date code.
  - Aged Products:** Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
  - Additional Requirements:** Sample Group A electrical tests are performed on a lot acceptance basis.

## Packaging

### 18 LEAD CERAMIC DIP



### 18 LEAD CERPACK



1. All dimensions in inches; millimeters are shown in parentheses.
2. All dimensions  $\pm .010$  ( $\pm 0.25$ mm) unless otherwise shown.

## Ordering Information

Example:

**HS 1 - 6514RH**

HARRIS CUSTOM INTEGRATED CIRCUITS DIVISION

PACKAGE

DEVICE

VERSION

### PACKAGE

FLAT PACK	9-
CERDIP	1-

**NOTICE:** Harris Semiconductor's products are sold by description only. Harris Semiconductor reserves the right to make changes in circuit design and/or specifications at any time without notice. Accordingly the reader is cautioned to verify that data sheets are current before placing orders.

## Sales Offices

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TWX: 510-959-6259

### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



**HARRIS**

CUSTOM INTEGRATED CIRCUITS DIVISION



# HS-6551RH

## 256 x 4 CMOS RAM

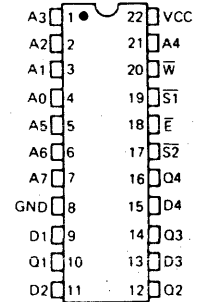
### Features

- FUNCTIONAL TOTAL DOSE
- LATCH-UP FREE TO
- LOW STANDBY POWER
- LOW OPERATING POWER
- FAST ACCESS TIME
- TTL COMPATIBLE IN/OUT
- HIGH OUTPUT DRIVE - 1TTL LOAD
- HIGH NOISE IMMUNITY
- ON CHIP ADDRESS REGISTER
- MILITARY TEMPERATURE RANGE
- THREE-STATE OUTPUTS
- 22 PIN PACKAGE FOR HIGH DENSITY

2 x 10<sup>4</sup> RAD Si  
 5.0 x 10<sup>11</sup> RAD Si/sec  
 550μW MAX  
 25mW/MHz MAX  
 300ns MAX

### Pinout

TOP VIEW



A – Address Input     $\bar{W}$  – Write Enable  
 $\bar{E}$  – Chip Enable    D – Data Input  
 $\bar{S}$  – Chip Select    Q – Data Output

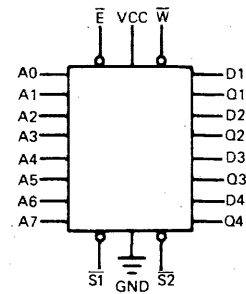
### Description

The HS-6551RH is a 256 by 4 static CMOS RAM fabricated using the Harris Programs Division radiation hardened self-aligned silicon gate technology. Synchronous circuit design techniques are employed to achieve high performance and low power operation.

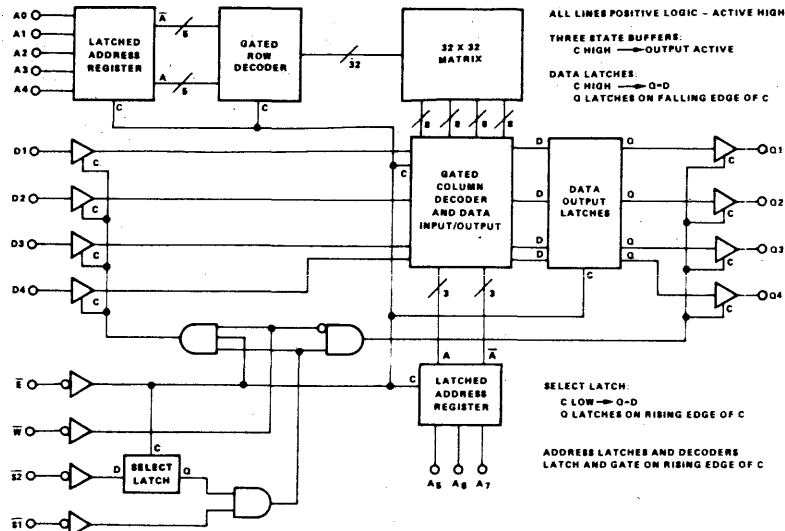
On-chip latches are provided for addresses, allowing efficient interfacing with microprocessor systems. The data output buffers can be forced to a high impedance state for use in expanded memory arrays.

The HS-6551RH is a fully static RAM and may be maintained in any state for an indefinite period of time.

### Logic Symbol



### Functional Diagram



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## Symbols and Abbreviations

This data sheet utilizes a new set of specification nomenclature. This new format is an IEEE and JEDEC supported standard for semiconductor memories. It is intended to clarify the symbols, abbreviations and definitions, and to make all memory data sheets consistent. We believe that, once acclimated, you will find this standardized format easy to read and use.

### ELECTRICAL PARAMETER ABBREVIATIONS

All abbreviations use upper case letters with no subscripts. The initial symbol is one of these four characters:

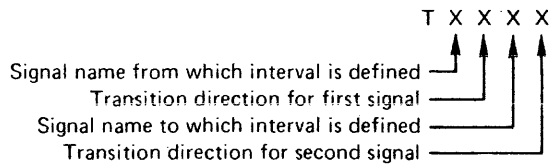
- V (Voltage)
- I (Current)
- P (Power)
- C (Capacitance)

The second letter specifies input (I) or output (O), and the third letter indicates the high (H), low (L) or off (Z) state of the pin during measurements. Examples:

- VIL — Input Low Voltage
- IOZ — Output Leakage Current

### TIMING PARAMETER ABBREVIATIONS

All timing abbreviations use upper case characters with no subscripts. The initial character is always T and is followed by four descriptors. These characters specify two signal points arranged in a "from-to" sequence that define a timing interval. The two descriptors for each signal point specify the signal name and the signal transitions. Thus the format is:



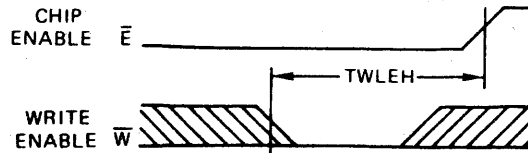
Signal Definitions:

- A = Address
- D = Data In
- Q = Data Out
- W = Write Enable
- E = Chip Enable
- S = Chip Select
- G = Output Enable

Transition Definitions:

- H = Transition to High
- L = Transition to Low
- V = Transition to Valid
- X = Transition to Invalid or Don't Care
- Z = Transition to Off (High Impedance)

EXAMPLE:



The example shows Write pulse setup time defined as TWLEH—Time from Write enable Low to chip Enable High.

### TIMING LIMITS

The table of timing values shows either a minimum or a maximum limit for each parameter. Input requirements are specified from the external system point of view. Thus, address set-up time is shown as a minimum since the system must supply at least that much time (even though most devices do not require it). On the other hand, responses from the memory are specified from the device point of view. Thus, the access time is shown as a maximum since the device never provides data later than that time.

### WAVEFORMS

WAVEFORM SYMBOL	INPUT	OUTPUT
	MUST BE VALID	WILL BE VALID
	CHANGE FROM H TO L	WILL CHANGE FROM H TO L
	CHANGE FROM L TO H	WILL CHANGE FROM L TO H
	DON'T CARE: ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
	—	HIGH IMPEDANCE

# Specifications HS-6551RH



ABSOLUTE MAXIMUM RATINGS			OPERATING RANGE	
Supply Voltage -VCC	+7V		Operating Supply Voltage -VCC	4.5V to 5.5V
Applied Input or Output Voltage	GND -0.3V VCC +0.3V		Operating Temperature	-55°C to +125°C
Storage Temperature	-65°C to +150°C			

*CAUTION: To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.*

## ELECTRICAL CHARACTERISTICS ⑤

D.C.

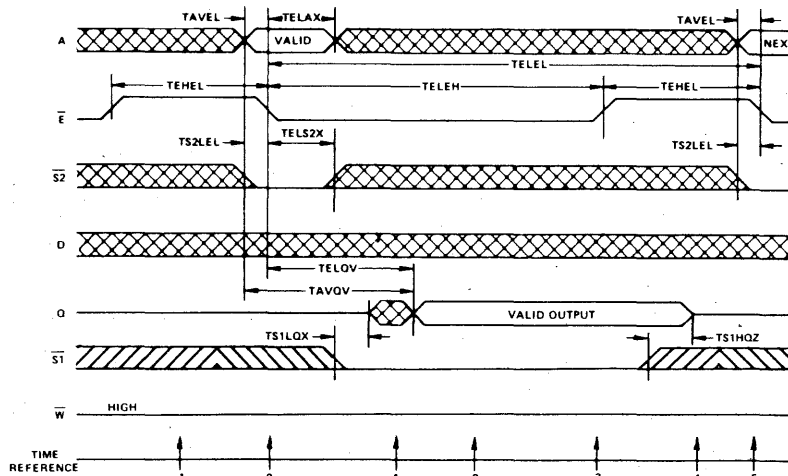
SYMBOL	PARAMETER	TEMP. & VCC = OPERATING RANGE		TEMP. = 25°C ① VCC = 5.0V			UNITS	TEST CONDITIONS
		MIN	MAX	MIN	TYP	MAX		
ICCSB	Standby Supply Current		100		10	100	μA	iO = 0 VI = VCC or GND
ICCOP	Operating Supply Current ②		4		1.5	2.5	mA	f = 1MHz, IO = 0 VI = VCC or GND
II	Input Leakage Current	-1.0	+1.0	-0.5	0.0	+0.5	μA	GND ≤ VI ≤ VCC
IOZ	Output Leakage Current	-1.0	+1.0	-0.5	0.0	+0.5	μA	GND ≤ VI ≤ VCC
VIL	Input Low Voltage	-0.3	0.8	-0.3		1.5	V	
VIH	Input High Voltage	VCC -2.0	VCC +0.3	2.5		5.3	V	
VOL	Output Low Voltage		0.4		0.2	0.35	V	IOL = 2.0mA
VOH	Output High Voltage	2.4		3.0	4.5		V	IOH = -1.0mA
CI	Input Capacitance ③		6		4	6	pF	VI = VCC or GND f = 1MHz
CO	Output Capacitance ③		10		6	10	pF	VI = VCC or GND f = 1MHz

A.C.

TELQV	Chip Enable Access Time		300		160	240	ns	④
TAVQV	Address Access Time		300		150	240	ns	④
TS1LQX	Chip Select 1 Output Enable Time		150		60	120	ns	④
TWLQZ	Write Enable Output Disable Time		150		60	120	ns	④
TS1HQZ	Chip Select 1 Output Disable Time		150		60	120	ns	④
TELEH	Chip Enable Pulse Negative Width	300		240	160		ns	④
TEHEL	Chip Enable Pulse Positive Width	150		70	50		ns	④
TAVEL	Address Setup Time	15		0	-10		ns	④
TS2LEL	Chip Select 2 Setup Time	15		0	-10		ns	④
TELAX	Address Hold Time	70		50	40		ns	④
TELS2X	Chip Select 2 Hold Time	70		50	40		ns	④
TDVWH	Data Setup Time	150		120	100		ns	④
TWHDX	Data Hold Time	0		0	0		ns	④
TWLS1H	Chip Select 1 Write Pulse Setup Time	180		150	120		ns	④
TWLEH	Chip Enable Write Pulse Setup Time	180		150	120		ns	④
TS1LWH	Chip Select 1 Write Pulse Hold Time	180		150	120		ns	④
TELWH	Chip Enable Write Pulse Hold Time	180		150	120		ns	④
TWLWH	Write Enable Pulse Width	180		150	120		ns	④
TELEL	Read or Write Cycle Time	450		270	170		ns	④

- NOTES:
1. All devices tested at worst case limits. Room temp., 5 volt data provided for information – not guaranteed.
  2. Operating Supply Current (ICCOP) is proportional to Operating Frequency. Example: Typical ICCOP = 1.5mA/MHz.
  3. Capacitance sampled and guaranteed – not 100% tested.
  4. AC Test Conditions: Inputs – TRISE = TFALL = 20nsec; Outputs – 1 TTL load and 50pF. All timing measurements at 1/2 VCC.
  5. Pre-radiation characteristics, see radiation effects for Post-Radiation characteristics.

### Read Cycle



TRUTH TABLE

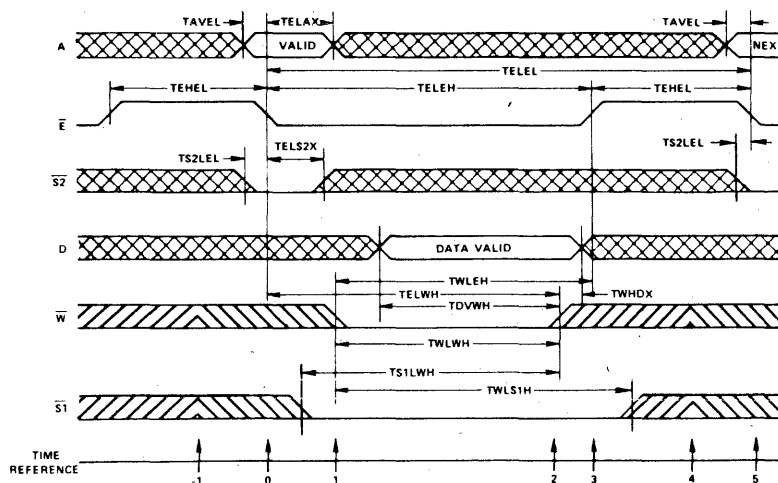
TIME REFERENCE	INPUTS				D	Q	FUNCTION
	$\bar{E}$	$\bar{S1}$	$\bar{S2}$	$\bar{W}$			
-1	H	H	X	X	X	Z	MEMORY DISABLED
0	X	L	H	V	X	Z	ADDRESSES AND $\bar{S2}$ ARE LATCHED, CYCLE BEGINS
1	L	L	X	H	X	X	OUTPUT ENABLED BUT UNDEFINED
2	L	L	X	H	X	V	DATA OUTPUT VALID
3	L	X	H	X	X	V	OUTPUTS LATCHES, VALID DATA
4	H	H	X	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	X	L	H	V	X	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The HS-6551RH Read Cycle is initiated by the falling edge of  $\bar{E}$ . This signal latches the input address word and  $\bar{S2}$  into on-chip registers providing that minimum setup and hold times are met. After the required hold time, these inputs may change state without affecting device operation.  $\bar{S2}$  acts as a higher order address and simplifies decoding. For the output to be read,  $\bar{E}$ ,  $\bar{S1}$  must be low and  $\bar{W}$  must be high.  $\bar{S2}$  must have been latched low on the falling

edge of  $\bar{E}$ . The output data will be valid at access time (TELQV).

The HS-6551RH has output data latches that are controlled by  $\bar{E}$ . On the rising edge of  $\bar{E}$  the present data is latched and remains in that state until  $\bar{E}$  falls. Either or both  $\bar{S1}$  or  $\bar{S2}$  may be used to force the output buffers into a high impedance state.

### Write Cycle



### TRUTH TABLE

TIME REFERENCE	INPUTS					OUTPUTS Q	FUNCTION
	E	S1	S2	W	A D		
-1	H	H	X	X	X	Z	MEMORY DISABLED
0	$\bar{X}$	X	L	X	V	X	CYCLE BEGINS, ADDRESSES AND $\bar{S2}$ ARE LATCHED
1	L	L	X	$\bar{X}$	X	X	WRITE PERIOD BEGINS
2	L	L	X	$\bar{X}$	X	V	DATA IN IS WRITTEN
3	$\bar{X}$	X	X	H	X	X	WRITE IS COMPLETED
4	H	H	X	X	X	X	PREPARE FOR NEXT CYCLE (SAME AS -1)
5	$\bar{X}$	X	L	X	V	X	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

In the Write Cycle the falling edge of  $\bar{E}$  latches the addresses and  $\bar{S2}$  into on-chip registers.  $\bar{S2}$  must be latched in the low state to enable the device. The write portion of the cycle is defined as  $\bar{E}$ ,  $\bar{W}$ ,  $\bar{S1}$  being low and  $\bar{S2}$  being latched low simultaneously. The  $\bar{W}$  line may go low at any time during the cycle providing that the write pulse setup times (TWLEH and TWLS1H) are met. The write portion of the cycle is terminated on the first rising edge of either  $\bar{E}$ ,  $\bar{W}$ , or  $\bar{S1}$ .

If a series of consecutive write cycles are to be executed, the  $\bar{W}$  line may be held low until all desired locations have been written. If this method is used, data setup and hold times must be referenced to the first rising edge of  $\bar{E}$  or  $\bar{S1}$ . By positioning the write pulse at different

times within the  $\bar{E}$  and  $\bar{S1}$  low time (TELEH) various types of write cycles may be performed. If the  $\bar{S1}$  low time (TS1LS1H) is greater than the  $\bar{W}$  pulse plus an output enable time (TS1LQX), a combination read-write cycle is executed. Data may be modified an indefinite number of times during any write cycle (TELEH).

The HS-6551RH may be used on a common I/O bus structure by tying the input and output pins together. The multiplexing is accomplished internally by the  $\bar{W}$  line. In the write cycle, when  $\bar{W}$  goes low, the output buffers are forced to a high impedance state. One output disable time delay (TWLQZ) must be allowed before applying input data to the bus.

### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $2 \times 10^4$  Rad Si (+10%) from a Gamma Cell 220 Cobalt 60 source or equivalent. The samples shall be biased at 5 volts with all inputs high. The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4) ICCSB at  $V_{CC} = 5$  volts will be measured and recorded for each device within one hour ( $\pm 15$  minutes) after irradiation. The lot will be accepted only if the average of these measured values is  $\leq 10$ mA.

### Radiation Effects

The HS-6551RH is a radiation hardened memory processed with the same mask set as is used for HARRIS' equivalent commercial part. Latchup free operation is achieved by the use of special starting material and improved total dose hardness is obtained with special high temperature processing cycles. These process techniques can, in principal, be applied to any standard HARRIS CMOS product.

The primary failure mode under exposure to ionizing radiation is an increase in static leakage current (ICCSB). Functional failure due to the increased leakage currents will typically occur for dose levels

in excess of  $5 \times 10^4$  RAD-Si. AC and DC parameters other than ICC will change less than 10% for total dose levels under  $5 \times 10^4$  RAD-Si. The excess leakage currents will anneal at room temperature and are typically reduced by a factor of 3-10 within 24 hours after irradiation.

On a production basis, HARRIS is able to perform screens only for total dose hardness. Transient radiation tests, however, have shown the following results:

Latchup free to doses  $\geq 5 \times 10^{11}$  rads/sec.  
Upset (loss of stored data) typically  $\geq 10^8$  rads/sec.

**HARRIS SEMICONDUCTOR PRODUCT FLOW**  
**MIL-M-38510/MIL-STD-883, METHOD 5004 CLASS B**

**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: (A) Fine (B) Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

Note:

Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.

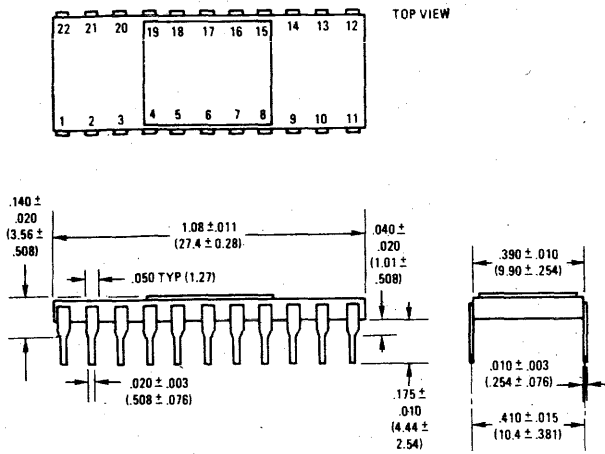
Branding: All devices are branded with the part number and EIA date code.

Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.

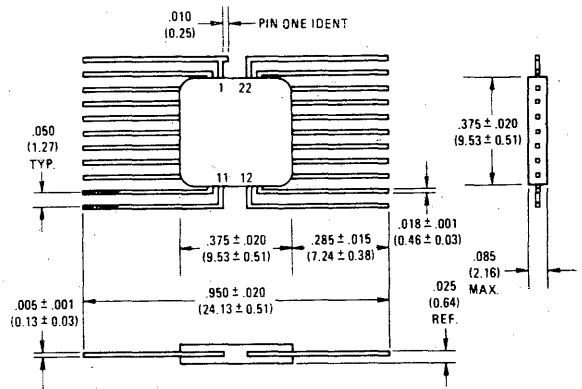
Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.

## Packaging

### 22 LEAD SIDE BRAZE DIP



### 22 LEAD FLATPACK



1. All dimensions in inches; millimeters are shown in parentheses.
2. All dimensions  $\pm .010$  ( $\pm 0.25$ mm) unless otherwise shown.

## Ordering Information

Example: **HS 1 - 6551 RH**

HARRIS PROGRAMS DIVISION — PACKAGE — DEVICE — VERSION

### PACKAGE

FLAT PACK	9-
CERDIP	1-

**NOTICE:** Harris Semiconductor's products are sold by description only. Harris Semiconductor reserves the right to make changes in circuit design and/or specifications at any time without notice. Accordingly, the reader is cautioned to verify that data sheets are current before placing orders.



# HARRIS

# HS-6564RH

## Radiation Resistant

## 8K x 8, 16K x 4 CMOS RAM

### Preliminary

### Features

- LOW POWER STANDBY
- LOW POWER OPERATION
- DATA RETENTION
- TTL COMPATIBLE IN/OUT
- THREE STATE OUTPUTS
- FAST ACCESS TIME
- MILITARY TEMPERATURE RANGE
- ON CHIP ADDRESS REGISTERS
- ORGANIZABLE 8Kx8 OR 16Kx4
- 40 PIN DIP PINOUT 2.000" x 0.900"
- FUNCTIONAL TOTAL DOSE
- DATA UPSET
- LATCH-UP FREE TO

500  $\mu$ W MAX  
180mW/MHz MAX  
3.0 V MIN

250 nsec TYP.  
-55°C TO +125°C

1x10<sup>5</sup> RAD Si  
>10<sup>6</sup> RAD Si/SEC  
>5x10<sup>7</sup> RAD Si/SEC

### Description

The HS-6564RH is a radiation resistant 64K bit, synchronous CMOS RAM. It consists of 16 HS-6504RH 4Kx1 radiation resistant CMOS RAMs, in leadless carriers, mounted on a ceramic substrate. The HS-6564RH is configured as an extra wide, standard length 40 pin DIP. The memory appears to the system as an array of 16 4Kx1 static 8Kx8 or 16Kx4 RAMs. The array is organized as two 8K by 4 blocks of RAM sharing only the address bus. The data inputs, data outputs, chip enables and write enables are separate for each block of RAM. This allows the user to organize the HS-6564RH RAM as either an 8K by 8 or a 16K by 4 array.

This 64K memory provides a unique blend of low power CMOS semiconductor technology and advanced packaging techniques. The HS-6564RH is intended for use in radiation environments where a large amount of RAM is needed, and where power consumption and board space are prime concerns. On-chip latches are provided for addresses, data input and data output allowing efficient interfacing with microprocessor systems. The data output can be forced to a high impedance for use in expanded memory arrays. The guaranteed low voltage data retention characteristics allow easy implementation of non-volatile read/write memory by using very small batteries mounted directly on the memory circuit board.

### Pinout

TOP VIEW

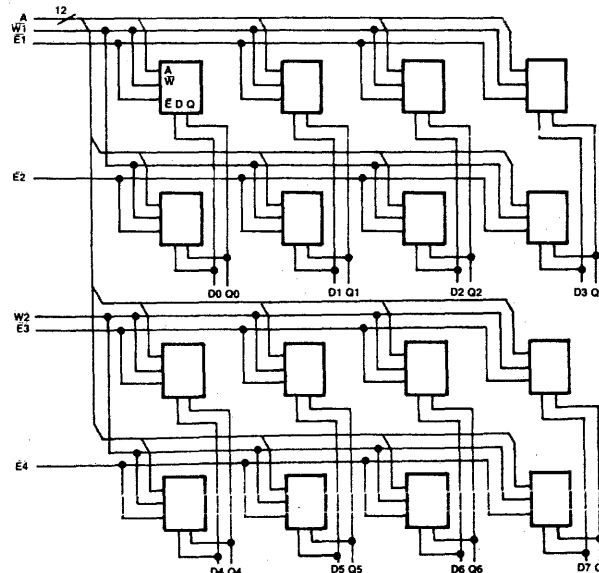
*GND	1	40	VCC*
Q4	2	39	Q0
D4	3	38	D0
Q5	4	37	Q1
D5	5	36	D1
A0	6	35	A6
A1	7	34	A7
A4	8	33	A8
E3	9	32	E1
*W2	10	31	W1
W2	11	30	W1*
E4	12	29	E2
A11	13	28	A3
A10	14	27	A2
A9	15	26	A5
D6	16	25	D2
Q6	17	24	Q2
D7	18	23	D3
Q7	19	22	Q3
*VCC	20	21	GND*

#### \*NOTES:

Pins 20 and 40 (VCC) are internally connected. Similarly pins 1 and 21 (Ground) are connected. The user is advised to connect all four VCC pins and Ground pins to his board busses. This will improve power distribution across the array and will enhance decoupling.

Pin 10 is internally connected to pin 11, and pin 30 is connected to pin 31. For those users wishing to preserve board compatibility with possible future RAM arrays, we recommend connections to the write lines be made at pins 11 and 31, leaving pins 10 and 30 free for future expansion.

### Functional Diagram



Information on this device is preliminary. Data is subject to change unless otherwise specifically agreed. No obligations are assumed for notice of change or future manufacture of this device.

CAUTION: These devices are sensitive to electronic discharge. Proper I.C. handling procedures should be followed.

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Harris Semiconductor  
CUSTOM/SEMICUSTOM

## Symbols and Abbreviations

This data sheet utilizes a new set of specification nomenclature. This new format is an IEEE and JEDEC supported standard for semiconductor memories. It is intended to clarify the symbols, abbreviations and definitions, and to make all memory data sheets consistent. We believe that, once acclimated, you will find this standardized format easy to read and use.

### ELECTRICAL PARAMETER ABBREVIATIONS

All abbreviations use upper case letters with no subscripts. The initial symbol is one of these four characters:

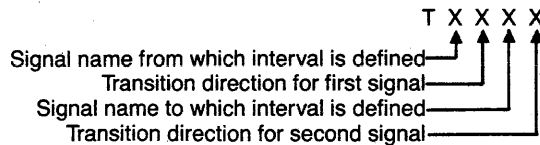
- V (Voltage)
- I (Current)
- P (Power)
- C (Capacitance)

The second letter specifies input (I) or output (O), and the third letter indicates the high (H), low (L) or off (Z) state of the pin during measurements. Examples:

- VIL – Input Low Voltage
- IOZ – Output Leakage Current

### TIMING PARAMETER ABBREVIATIONS

All timing abbreviations use upper case characters with no subscripts. The initial character is always T and is followed by four descriptors. These characters specify two signal points arranged in a "from-to" sequence that define a timing interval. The two descriptors for each signal point specify the signal name and the signal transitions. Thus the format is:



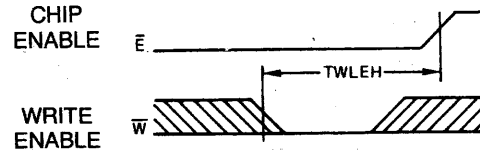
Signal Definitions:

- A = Address
- D = Data In
- Q = Data Out
- W = Write Enable
- E = Chip Enable
- S = Chip Select
- G = Output Enable

Transition Definitions:

- H = Transition to High
- L = Transition to Low
- V = Transition to Valid
- X = Transition to Invalid or Don't Care
- Z = Transition to Off (High Impedance)

### EXAMPLE:



The example shows Write pulse setup time defined as TWLEH-Time from Write enable Low to chip Enable High.

### TIMING LIMITS

The table of timing values shows either a minimum or a maximum limit for each parameter. Input requirements are specified from the external system point of view. Thus, address set-up time is shown as a minimum since the system must supply at least that much time (even though most devices do not require it). On the other hand, responses from the memory are specified from the device point of view. Thus, the access time is shown as a maximum since the device never provides data later than that time.

### WAVEFORMS

WAVEFORM SYMBOL	INPUT	OUTPUT
—	MUST BE VALID	WILL BE VALID
	CHANGE FROM H TO L	WILL CHANGE FROM H TO L
	CHANGE FROM L TO H	WILL CHANGE FROM L TO H
	DON'T CARE: ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
	—	HIGH IMPEDANCE





# Specifications HS-6564RH

## ABSOLUTE MAXIMUM RATINGS\*

Supply Voltage - (VCC - GND)	-0.3V to +7.0V
Input or Output Voltage Applied	(GND - 0.3V) to (VCC + 0.3V)
Storage Temperature	-65°C to +150°C

## OPERATING RANGE

Operating Supply Voltage	+4.5 to +5.5V
Operating Temperature	-55°C to +125°C

\* CAUTION: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and functional operation of the device at these or at any other conditions above those indicated in the operational sections of this specification is not implied.

## ELECTRICAL CHARACTERISTICS ④

SYMBOL	PARAMETER	TEMP. & VCC = OPERATING RANGE		TYPICAL TEMP. = 25°C ① VCC = 5.0V		UNITS	TEST CONDITIONS
		MIN	MAX	PRE RAD	POST RAD		
ICCSB	Standby Supply Current		800	80	80	μA	IO = 0 VI = VCC or GND
ICCOP1	Operating Supply Current (8K × 8) ②		56	36		mA	f = 1MHz, IO = 0 VI = VCC or GND
ICCOP2	Operating Supply Current (16K × 4) ②		28	18		mA	f = 1MHz, IO = 0 VI = VCC or GND
ICCDR	Data Retention Supply Current		600	48	48	μA	IO = 0, VCC = 3.0 VI = VCC or GND
VCCDR	Data Retention Supply Voltage	3.0		2.6	1.8	V	
IIA	Address Input Leakage	-20	+20			μA	GND ≤ VI ≤ VCC
IID1	Data Input Leakage (8K × 8)	-3	+3			μA	GND ≤ VI ≤ VCC
IID2	Data Input Leakage (16K × 4)	-5	+5			μA	GND ≤ VI ≤ VCC
IIE1	Enable Input Leakage (8K × 8)	-10	+10			μA	GND ≤ VI ≤ VCC
IIE2	Enable Input Leakage (16K × 4)	-5	+5			μA	GND ≤ VI ≤ VCC
IIW	Write Enable Input Leakage (Each)	-10	+10			μA	GND ≤ VI ≤ VCC
IOZ1	Output Leakage (8K × 8)	-20	+20	±4	±4	μA	GND ≤ VO ≤ VCC
IOZ2	Output Leakage (16K × 4)	-40	+40	±8	±8	μA	GND ≤ VO ≤ VCC
VIL	Input Low Voltage	-0.3	0.8	1.5	1.3	V	
VIH1	Input High Level (Except E and W)	VCC - 1.5	VCC + 0.3	2.7	2.5	V	
VIH2	Input High Level (E and W)	VCC - 1.0	VCC + 0.3	2.9	2.1	V	
VOL	Output Low Voltage		0.4	0.2	0.15	V	IO = 2.0mA
VOH	Output High Voltage	2.4		4.6	4.0	V	IO = -1.0mA
CIA	Address Input Capacitance ③		200			pF	f = 1MHz, VI = VCC or GND
CID1	Data Input Capacitance (8K × 8) ③		50			pF	f = 1MHz, VI = VCC or GND
CID2	Data Input Capacitance (16K × 4) ③		100			pF	f = 1MHz, VI = VCC or GND
CIE1	Enable Input Capacitance (8K × 8) ③		160			pF	f = 1MHz, VI = VCC or GND
CIE2	Enable Input Capacitance (16K × 4) ③		80			pF	f = 1MHz, VI = VCC or GND
CIW	Write Enable Input Capacitance (Each) ③		100			pF	f = 1MHz, VI = VCC or GND
CO1	Output Capacitance (8K × 8) ③		50			pF	f = 1MHz, VO = VCC or GND
CO2	Output Capacitance (16K × 4) ③		100			pF	f = 1MHz, VO = VCC or GND

### NOTES:

① Each individual RAM in the leadless carrier is fully tested at worst case limits of temperature and voltage. The complete assembled HS-6564RH array is tested at room temperature only. The worst case parameters are guaranteed over the specified temperature and voltage ranges. Room temperature, 5 volt data is provided for information purposes and is not guaranteed.

② Operating supply current is proportional to operating frequency. ICCOP is specified at an operating frequency of 1MHz, indicating repetitive accessing at a 1μs rate. Operation at slower rates will decrease ICCOP proportionally.

③ Capacitance sampled and guaranteed - not 100% tested.

④ Pre Radiation and Post Radiation limits.



ELECTRICAL CHARACTERISTICS

A.C.

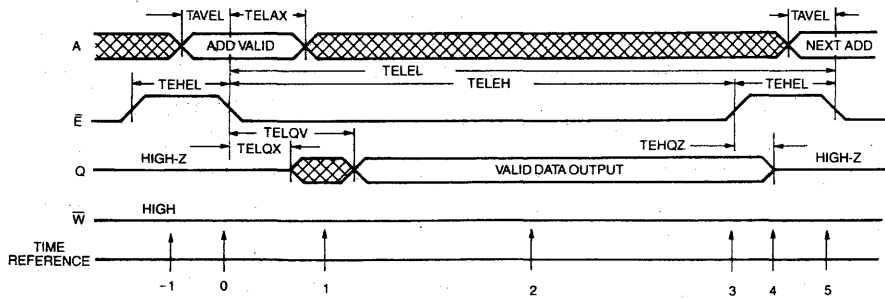
SYMBOL	PARAMETER	TEMP. & VCC = OPERATING RANGE		TYPICAL TEMP. = 25°C VCC = 5.0V		UNITS	TEST CONDITIONS
		MIN	MAX	PRE RAD	POST RAD		
TELQV	Chip Enable Access		350	200	230	ns	⑤
TAVQV	Address Access (TAVQV = TELQV + TAVEL)		400	220	260	ns	
TELOX	Output Enable	20	120	70	80	ns	
TEHQZ	Output Disable		130	80	90	ns	
TELEL	Read or Write Cycle	480		250	320	ns	
TELEH	Chip Enable Low	350		200	230	ns	
TEHEL	Chip Enable High	130		50	90	ns	
TAVEL	Address Setup	50		20	30	ns	
TELAX	Address Hold	50		30	35	ns	
TWLWH	Write Enable Low	150		90	100	ns	
TWLEH	Write Enable Setup	250		190	190	ns	
TWLEL	Early Write Setup (Write Mode)	10		-5	0	ns	
TWHEL	Write Enable Read Setup	10		-5	0	ns	
TELWX	Early Write Hold (Write Mode)	100				ns	
TDVWL	Data Setup	10		-5	20	ns	
TDVEL	Early Write Data Setup	10		-5	20	ns	
TWLDX	Data Hold	100		70	80	ns	
TELDX	Early Write Data Hold	100		70	80	ns	
TQVWL	Data Valid to Write (Read-Modify-Write)	0		0	0	ns	

NOTES:

⑤ AC Test Conditions:

Inputs - Trise = Tfall = 20ns.  
Outputs - CLOAD = 100pF.  
Timing measured at 1.5V reference level.

Read Cycle



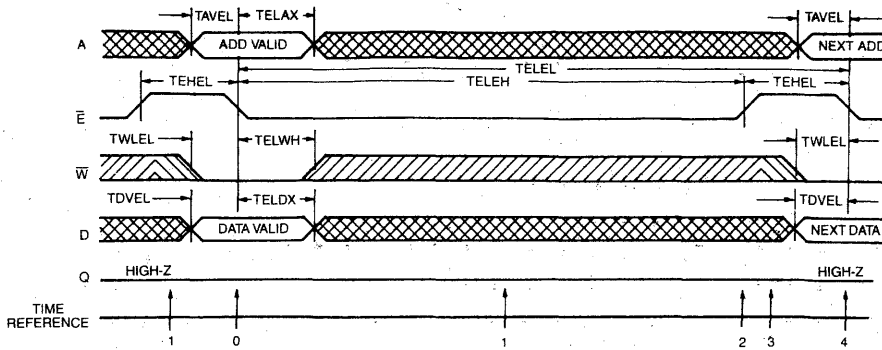
TRUTH TABLE

TIME REFERENCE	INPUTS			OUTPUT Q	FUNCTION
	E	W	A		
-1	H	X	X	Z	MEMORY DISABLED
0		H	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	H	X	X	OUTPUT ENABLED
2	L	H	X	V	OUTPUT VALID
3		H	X	V	READ ACCOMPLISHED
4	H	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
5		H	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The address information is latched in the on chip registers on the falling edge of  $\bar{E}$  (T = 0). Minimum address set up and hold time requirements must be met. After the required hold time, the addresses may change state without affecting device operation. During time (T = 1) the output becomes enabled

but data is not valid until during time (T = 2).  $\bar{W}$  must remain high until after time (T = 2). After the output data has been read,  $\bar{E}$  may return high (T = 3). This will disable the output buffer and ready the RAM for the next memory cycle (T = 4).

### Early Write Cycle



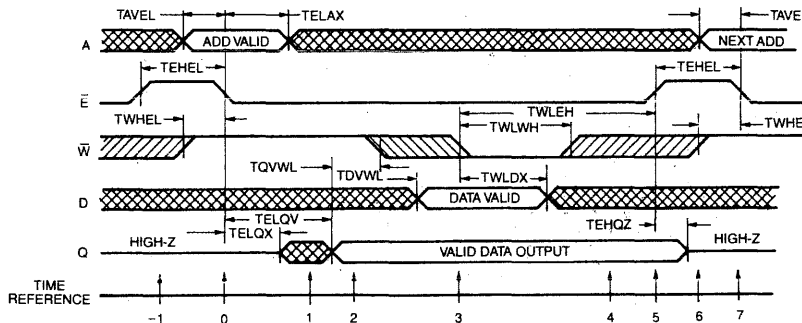
**TRUTH TABLE**

TIME REFERENCE	$\bar{E}$	W	A	D	OUTPUT Q	FUNCTION
-1	H	X	X	X	Z	MEMORY DISABLED
0	L	X	V	V	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	X	X	X	Z	WRITE IN PROGRESS INTERNALLY
2	L	X	X	X	Z	WRITE COMPLETED
3	H	X	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
4	L	V	V	V	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The early write cycle is the only cycle where the output is guaranteed not to become active. On the falling edge of  $\bar{E}$  (T = 0), the addresses, the write signal, and the data input are latched in on chip registers. The logic value of  $\bar{W}$  at the time  $\bar{E}$  falls determines the state of the output buffer for that cycle. Since  $\bar{W}$  is low when  $\bar{E}$  falls, the output buffer is latched into the

high impedance state and will remain in that state until  $\bar{E}$  returns high (T = 2). For this cycle, the data input is latched by  $\bar{E}$  going low; therefore data set up and hold times should be referenced to  $\bar{E}$ . When  $\bar{E}$  (T = 2) returns to the high state, the output buffer disables and all signals are unlatched. The device is now ready for the next cycle.

### Read Modify Write Cycle



**TRUTH TABLE**

TIME REFERENCE	$\bar{E}$	W	A	D	OUTPUT Q	FUNCTION
-1	H	X	X	X	Z	MEMORY DISABLED
0	L	H	V	X	Z	CYCLE BEGINS, ADDRESSES ARE LATCHED
1	L	H	X	X	X	OUTPUT ENABLED
2	L	H	X	X	V	OUTPUT VALID, READ AND MODIFY TIME
3	L	L	X	V	V	WRITE BEGINS, DATA IS LATCHED
4	L	X	X	X	V	WRITE IN PROGRESS INTERNALLY
5	L	X	X	X	V	WRITE COMPLETED
6	H	X	X	X	Z	PREPARE FOR NEXT CYCLE (SAME AS -1)
7	L	H	V	X	Z	CYCLE ENDS, NEXT CYCLE BEGINS (SAME AS 0)

The read modify write cycle begins as all other cycles on the falling edge of  $\bar{E}$  (T = 0). The  $\bar{W}$  line should be high at (T = 0) in order to latch the output buffers in the active state. During (T = 1) the output will be active but not valid until (T = 2). On the falling edge of the  $\bar{W}$  (T = 3) the data present at the output and input are latched. The  $\bar{W}$  signal also latches itself on its low

going edge. All input signals excluding  $\bar{E}$  have been latched and have no further effect on the RAM. The rising edge of  $\bar{E}$  (T = 5) completes the write portion of the cycle and unlatches all inputs and output. The output goes to a high impedance and the RAM is ready for the next cycle.

**NOTES:** In the above descriptions the numbers in parenthesis (T = n) refer to the respective timing diagrams. The numbers are located on the time reference line below each diagram. The timing diagrams shown are only examples and are not the only valid method of operation.



### Board Size Tradeoffs

Printed circuit board real estate is a costly commodity. Actual board costs depend on layout tolerances, density, complexity, number of layers, choice of board material, and other factors.

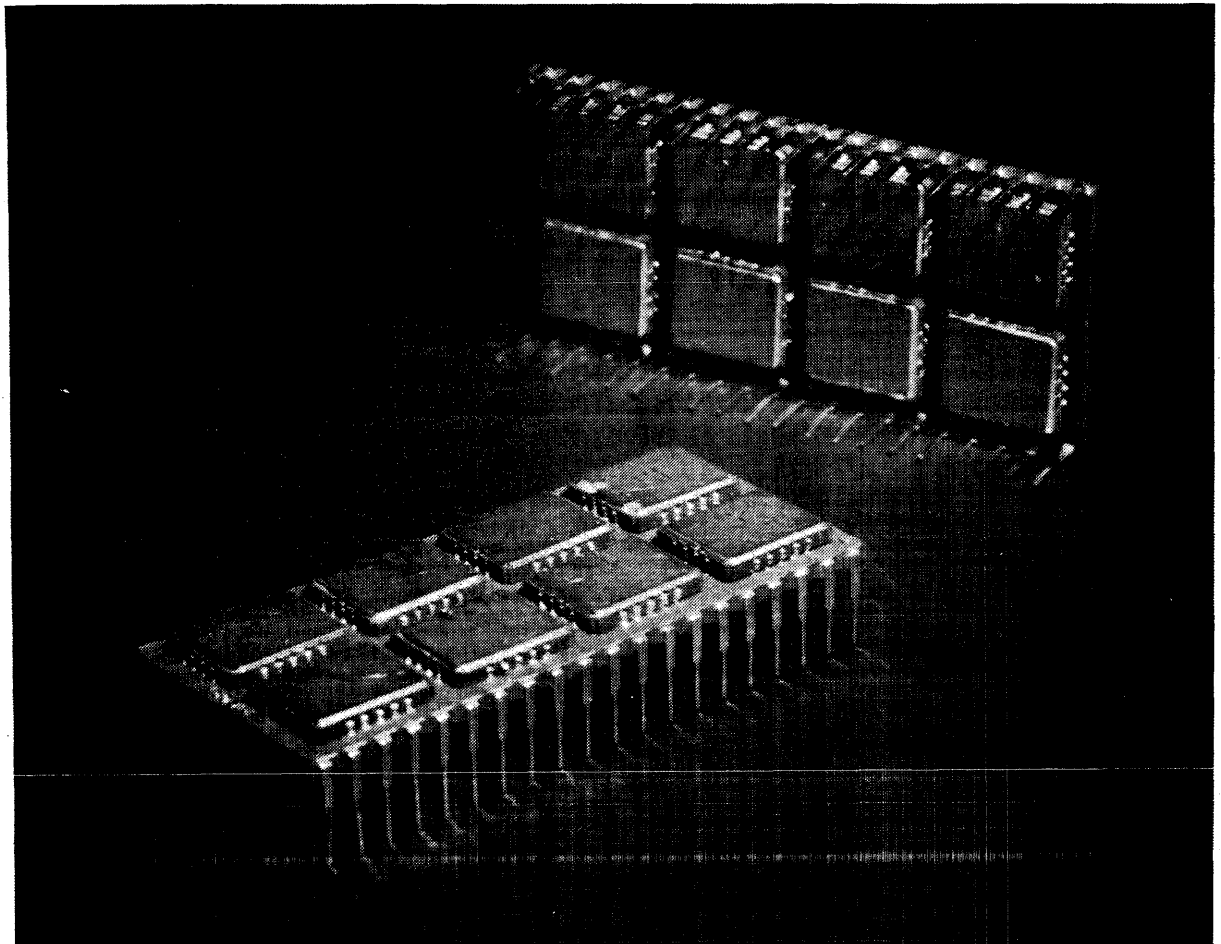
The following table compares board space for 16 standard DIP 4K RAMs to the HS-6564RH RAM array. Both fine line, close tolerance layout and standard "easy" layout board sizes are shown in the comparison.

#### 64K ARRAY OF 16 4K RAMS ON A PC BOARD V.S. THE HS-6564RH

PACKAGE	CIRCUIT SUBSTRATE	SIZE
18 Pin DIP	Standard Two Sided PCB	12 to 15 sq. in.
18 Pin DIP	Fine Line or Multilayer PCB	9 to 11 sq. in.
18 Pin Leadless Carrier	Multilayer Alumina Substrate	3 to 5 sq. in.
HS-6564RH	Two Sided Mounting Multilayer Alumina Substrate	2 sq. in.

The cost of semiconductor circuits decline with time. If actual costs were included, they would be out of date in a very short time. We urge you to contact your local Harris office or sales representative for accurate pricing allowing cost tradeoff

analysis. In your cost analysis, also consider the advantages of a lighter, smaller overall package for your system. Consider how much more valuable your system will be when the memory array size is decreased to about 1/6 of normal size.



HS-6564RH - 64K BIT CMOS RAM

HARRIS MIL-100

SC

- Internal V
- Stabilizat
- Tempera
- Constant
- Seal: F
- Initial Ele
- Burn-In T
- Final Ele
- 100% go
- External \

IS PERFORMED

ices are assigne  
ices are brande  
t that has been f  
tion requirement

e Group A electr.

Harris Semiconductor

CUSTOM/SEMICUSTOM

**To Organize 8K x 8:**

Connect: E1 with E3  
E2 with E4  
W1 with W2

(Pin  
(Pin  
(Pin

**Concerns for Proper Operation of Chip**

The transition between blocks of RAM requires the chip enables as if there were only two, E1 and E2. Between chip enables must be treated with high at least one chip enable high time (TEH) devices whose outputs are tied common.

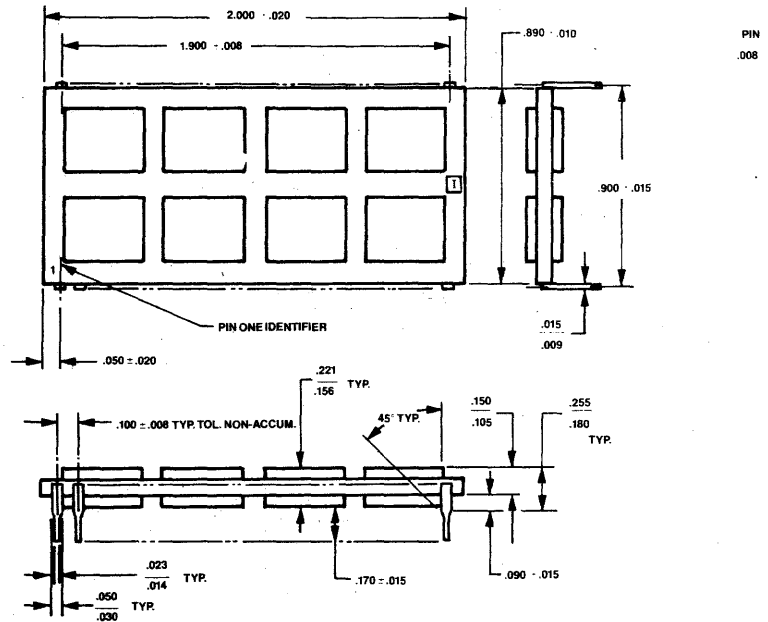
**Printed Circuit Board Mounting:**

The leadless chip carrier packages used in this design are mounted to VCC or GND. The designer should be aware that the chip must be pressed completely down against the surface of the PCB to help prevent the leadless carriers from moving.

HARRIS CMOS RAMs are designed to operate over a wide temperature range and are guaranteed over temperature. The following are the minimum operating conditions:

1. Chip Enable ( $\bar{E}$ ) must be held high.
2. All other inputs should be held high.
3. Inputs which are to be held high must be pulled up and power down transition time must be limited.
4. The RAM can begin operation when the chip enable is held high.

**Packaging**



**NOTES:**

1. ALL DIMENSIONS ARE IN INCHES UNLESS OTHERWISE SPECIFIED.
2. FLAT
3. DIE AREA

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**CUSTOM INTEGRATED CIRCUITS I**

VCC

E1



### Board Size Tradeoffs

Printed circuit board real estate is a costly commodity. Actual board costs depend on layout tolerances, density, complexity, number of layers, choice of board material, and other factors.

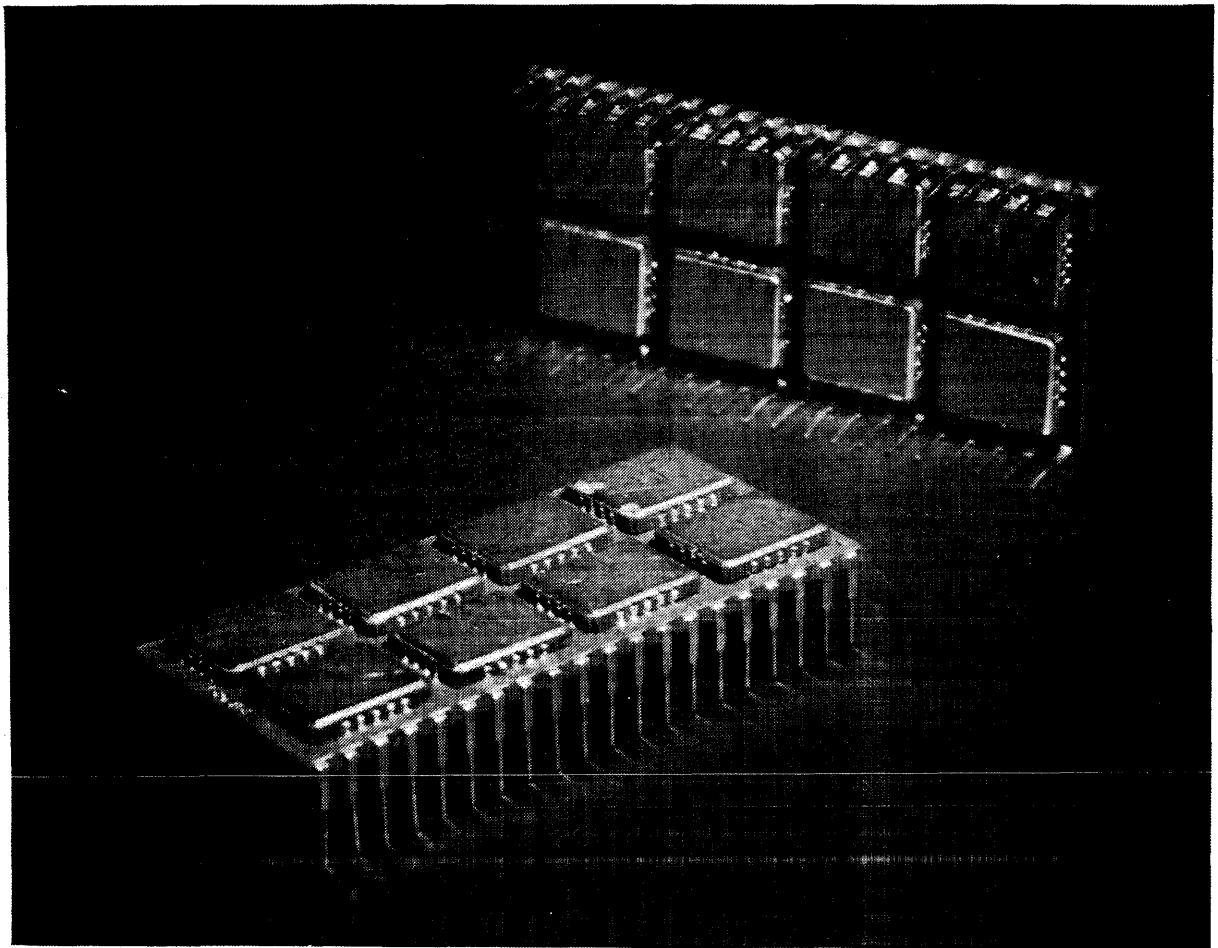
The following table compares board space for 16 standard DIP 4K RAMs to the HS-6564RH RAM array. Both fine line, close tolerance layout and standard "easy" layout board sizes are shown in the comparison.

#### 64K ARRAY OF 16 4K RAMs ON A PC BOARD V.S. THE HS-6564RH

PACKAGE	CIRCUIT SUBSTRATE	SIZE
18 Pin DIP	Standard Two Sided PCB	12 to 15 sq. in.
18 Pin DIP	Fine Line or Multilayer PCB	9 to 11 sq. in.
18 Pin Leadless Carrier	Multilayer Alumina Substrate	3 to 5 sq. in.
HS-6564RH	Two Sided Mounting Multilayer Alumina Substrate	2 sq. in.

The cost of semiconductor circuits decline with time. If actual costs were included, they would be out of date in a very short time. We urge you to contact your local Harris office or sales representative for accurate pricing allowing cost tradeoff

analysis. In your cost analysis, also consider the advantages of a lighter, smaller overall package for your system. Consider how much more valuable your system will be when the memory array size is decreased to about 1/6 of normal size.



HS-6564RH - 64K BIT CMOS RAM

## Organization Guide

### To Organize 8K × 8:

Connect:  $\bar{E}1$  with  $\bar{E}3$   
 $\bar{E}2$  with  $\bar{E}4$   
 $\bar{W}1$  with  $\bar{W}2$

(Pins 9 + 32)  
(Pins 12 + 29)  
(Pins 11 + 31)

### To Organize 16K × 4:

Connect: Q0 with Q4  
D0 with D4  
Q1 with Q5  
D1 with D5  
D2 with D6  
Q2 with Q6  
D3 with D7  
Q3 with Q7

Optional  $\bar{W}1$  may be common with  $\bar{W}2$

(Pins 2 + 39)  
(Pins 3 + 38)  
(Pins 4 + 37)  
(Pins 5 + 36)  
(Pins 16 + 25)  
(Pins 17 + 24)  
(Pins 18 + 23)  
(Pins 19 + 22)  
(Pins 11 + 31)

### Concerns for Proper Operation of Chip Enables:

The transition between blocks of RAM requires a change in the chip enable being used. When operating in the 8K × 8 mode, use the chip enables as if there were only two,  $\bar{E}1$  and  $\bar{E}2$ . In the 16K × 4 mode, all chip enables must be treated separately. Transitions between chip enables must be treated with the same timing constraints that apply to any one chip enable. All chip enables must be high at least one chip enable high time (TEHEL) before any chip enable can fall. More than one chip enable low simultaneously, for devices whose outputs are tied common either internally or externally, is an illegal input condition and must be avoided.

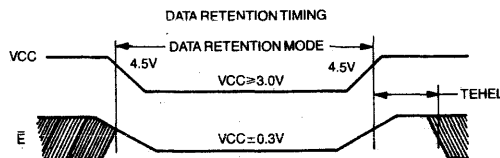
### Printed Circuit Board Mounting:

The leadless chip carrier packages used in the HS-6564RH have conductive lids. These lids are electrically floating, not connected to VCC or GND. The designer should be aware of the possibility that the carriers on the bottom side could short conductors below if pressed completely down against the surface of the circuit board. The pins on the package are designed with a standoff feature to help prevent the leadless carriers from touching the circuit board surface.

## Low Voltage Data Retention

HARRIS CMOS RAMs are designed with battery backup in mind. Data retention voltage and supply current are guaranteed over temperature. The following rules insure data retention:

1. Chip Enable ( $\bar{E}$ ) must be held high during data retention; within  $VCC + 0.3V$  to  $VCC - 0.3V$ .
2. All other inputs should be held either high (at CMOS VCC) or at ground to minimize ICCDR.
3. Inputs which are to be held high (e.g.  $\bar{E}$ ) must be kept between  $VCC + 0.3V$  and 70% of VCC during the power up and power down transitions.
4. The RAM can begin operation one TEHEL after VCC reaches the minimum operating voltage (4.5 volts).





### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose
- (4) The samples will be tested to the data sheet limits within one hour ( $\pm 15$  minutes) after irradiation. The lot will be accepted only if all units, exclusive of non-radiation failures, meet the data sheet limits.

Radiation level of  $1 \times 10^5$  Rad Si ( $\pm 10\%$ ) from a Gamma Cell 220 Cobalt 60 source or equivalent. The samples shall be biased at 5 volts with all inputs high. The dose rate shall be between 50 rads/sec and 200 rads/sec.

### Radiation Effects

The HS-6564RH is a radiation resistant memory module designed to survive in a radiation environment and to meet the electrical characteristics and be pin compatible to the Harris equivalent commercial part. Latchup free operation, achieved by the use of special starting material and improved total dose hardness, is obtained with special high temperature processing cycles. These process techniques can, in principle, be applied to any standard HARRIS CMOS product.

On a production basis, HARRIS only performs screens for total dose hardness to a level of  $1 \times 10^5$  rad-Si. Transient radiation tests, however, have shown the following results:

Latchup free to doses  $\geq 5 \times 10^{11}$  rads/sec.  
Upset (loss of stored data) typically  $\geq 10^8$  rads/sec.

### Test Product Flow

**HARRIS SEMICONDUCTOR PRODUCT FLOW  
MIL-STD-883, METHOD 5004 CLASS B  
100% SCREENING PROCEDURE\***

	SCREEN	MIL-STD-883 METHOD/COND.
①	Internal Visual	2010 Cond. B.
②	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
③	Temperature Cycling	1010 Cond. C
④	Constant Acceleration	2001 Cond. E; Y1 plane
⑤	Seal: Fine Gross	1014 Cond. A or B 1014 Cond. C2
⑥	Initial Electrical	Harris Specifications
⑦	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent)
⑧	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
⑨	External Visual	2009 Sample Inspection

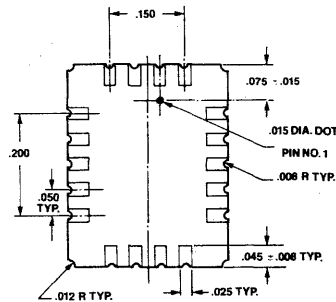
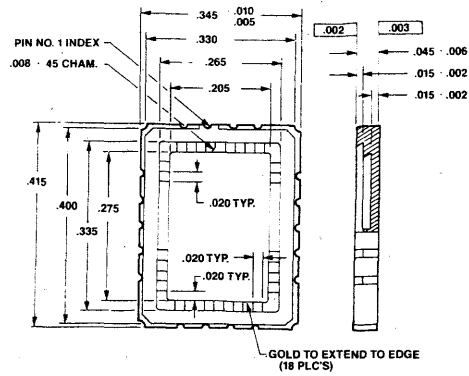
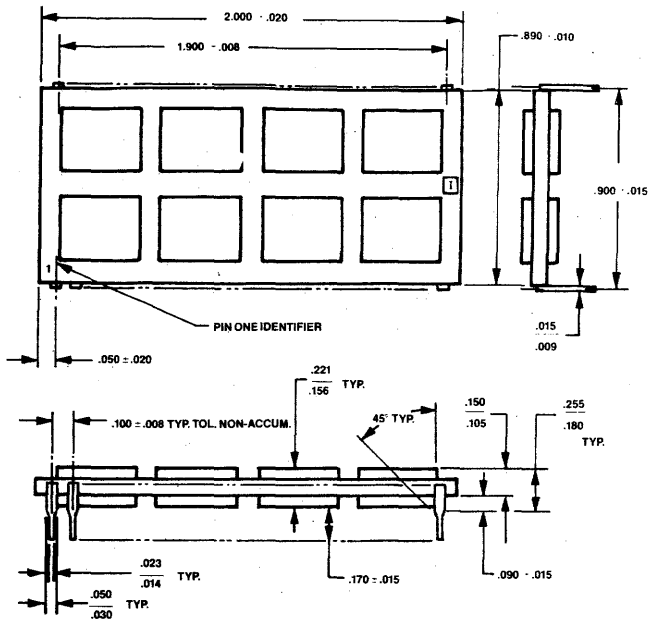
\* ALL SCREENING IS PERFORMED AT LEADLESS CARRIER LEVEL EXCEPT FINAL ELECTRICAL

#### NOTE:

- Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.
- Branding: All devices are branded with the part number and EIA date code.
- Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
- Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.



## Packaging



### NOTES:

1. ALL EXPOSED METALLIZED AREAS SHALL BE GOLD PLATED 50 MICRO INCHES MIN. THICKNESS OVER NICKEL PLATE.
2. FLATNESS PERTAINS TO METALLIZED PADS ONLY.
3. DIE ATTACH PAD TO BE ELECTRICALLY CONNECTED TO PIN NO. 18.

*NOTICE: Harris Semiconductor's products are sold by description only. Harris Semiconductor reserves the right to make changes in circuit design and/or specifications at any time without notice. Accordingly the reader is cautioned to verify that data sheets are current before placing orders.*

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P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



# HARRIS

CUSTOM INTEGRATED CIRCUITS DIVISION



Preliminary

# HS-3516RH

High Slew Rate, Wideband,  
Radiation Resistant,  
Operational Amplifier

## Features

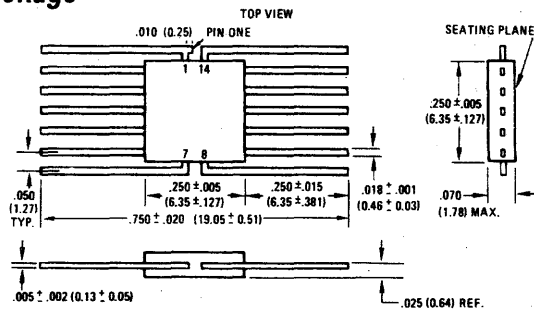
- HIGH SLEW RATE . . . . .  $> \pm 22V/\mu s$
- FAST SETTLING TIME . . . . . 450ns
- WIDE POWER BANDWIDTH . . . . . 12MHz
- LOW OFFSET VOLTAGE . . . . .  $\pm 5mV$
- LOW POWER SUPPLY CURRENT . . . . . 6.5mA
- SHORT CIRCUIT PROTECTION
- RADIATION ENVIRONMENT
  - NEUTRON FLUENCE ( $\phi$ )  $5 \times 10^{12} n/cm^2$  ( $E \geq 10KeV$ )
  - GAMMA RATE ( $\dot{\gamma}$ ) . . . . .  $1 \times 10^9$  RADs Si/s
  - GAMMA DOSE ( $\gamma$ ) . . . . .  $1 \times 10^6$  RADs Si

## Description

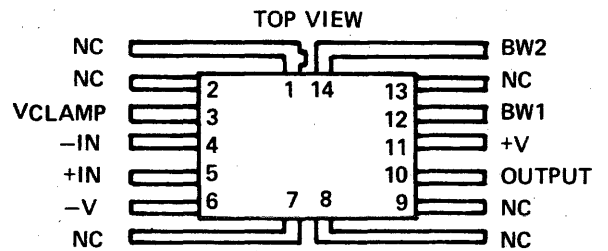
The HS-3516RH is a monolithic, high slew rate, wideband, radiation resistant, operational amplifier. It provides a bandwidth (unity gain stable) of greater than 10MHz and a slew rate in excess of 22V/ $\mu$ sec. Optional frequency compensation adjustment is provided. The HS-3516RH has an internal unity gain frequency compensation capacitor which is internally connected. A clamp node feature enables the user to clamp the output voltage via pin 3 which can source or sink up to 3mA for high frequency clamped switching purposes.

This device is designed to operate from  $-55^\circ C$  to  $+125^\circ C$  and in strategic-level radiation environments.

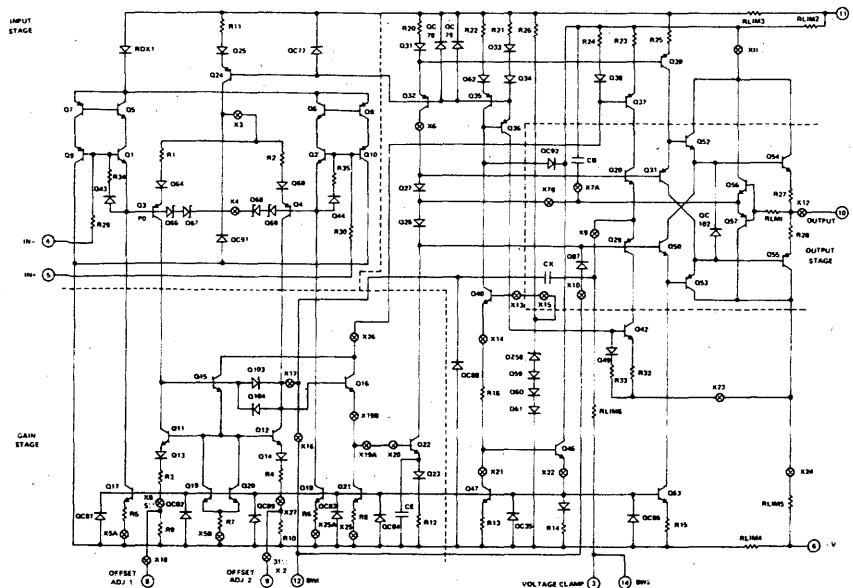
## Package



## Pinout



## Schematic



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Harris Semiconductor

CUSTOM/SEMICUSTOM



# Specifications HS-3516RH

## ABSOLUTE MAXIMUM RATINGS

Voltage Between +V and -V terminals	40V	Internal Power Dissipation	625mW
Differential Input Voltage	±15V	Storage Temperature Range	-65°C to +175°C
Output Short Circuit Duration	Indefinite	Operating Temperature Range	-55°C to +125°C

**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

## ELECTRICAL CHARACTERISTICS +V = 15V, -V = -15V, T<sub>A</sub> = -55°C to +125°C

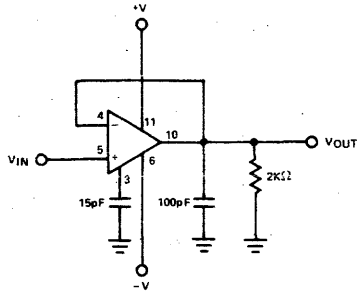
PARAMETER	TEMP	MIN	TYP	MAX	UNITS
Offset Voltage	25°C		1	3	mV
	125°C		2	5	mV
	-55°C		3	6	mV
Input Bias Current	25°C		35	100	nA
	125°C		50	100	nA
	-55°C			200	nA
Input Offset Current	25°C		30	100	nA
	125°C			150	nA
	-55°C			250	nA
Input Resistance	25°C		>100		MΩ
Large Signal Voltage Gain <sup>1</sup>	25°C	90			dB
	Full	90			dB
Common Mode Rejection Ratio <sup>2</sup>	25°C	80	115		dB
	Full	80	90		dB
Supply Current	25°C		5.0	6.5	mA
	125°C			6.5	mA
	-55°C			8.7	mA
Power Supply Rejection Ratio <sup>3</sup>	25°C	80			dB
	Full	80			dB
Output Voltage Swing	25°C	+12.5	+13.0		V
		-11.0	-12.0		V
	125°C	+12.5	+13.0		V
		-11.0	-12.0		V
	-55°C	+12.0			V
		-11.0			V
Output Short Circuit Current	25°C		27	45	mA
	125°C			45	mA
	-55°C			60	mA
Gain Bandwidth Product	25°C		12.0		MHz
	Full		10.0		MHz
Slew Rate <sup>3, 5</sup>	25°C	22	25		V/μs
Rise Time <sup>4</sup>	25°C		30	35	ns
Overshoot <sup>4</sup>	25°C		15	35	%
Overdrive Recovery Time	25°C		2	5	μs
Settling Time <sup>4</sup>	25°C		180	450	ns
	Full			450	ns
Output Clamp Voltage	25°C			0.4	V
Input Clamp (+) I <sub>cn-</sub> Current (pin 3)	25°C	-1.0		-3.3	mA
	125°C	-1.0		-3.3	mA
	-55°C	-0.8		-3.5	mA
Input Clamp (-) I <sub>cn+</sub> Current (pin 3)	25°C	+0.5		+3.0	mA
	125°C	+0.5		+3.0	mA
	-55°C	+0.3		+3.2	mA

### NOTES:

- V<sub>O</sub> = ±10V, R<sub>L</sub> = 2K
- V<sub>cm</sub> = ±10V, R<sub>L</sub> = 2K
- ΔV = ±5.0V
- A<sub>V</sub> = +1, V<sub>IN</sub> = 1V, R<sub>L</sub> = 2K, C<sub>L</sub> = 100pf
- C<sub>BW</sub> = 0, V<sub>O</sub> = ±10V, R<sub>L</sub> = 2K, C<sub>L</sub> = 100pf

## Test Circuits

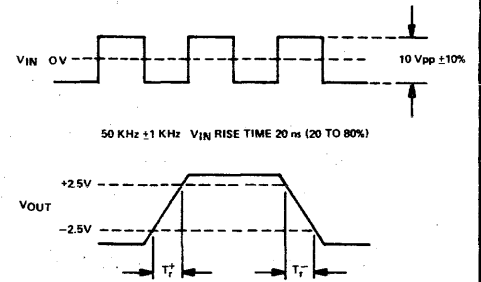
### SLEW RATE CIRCUIT



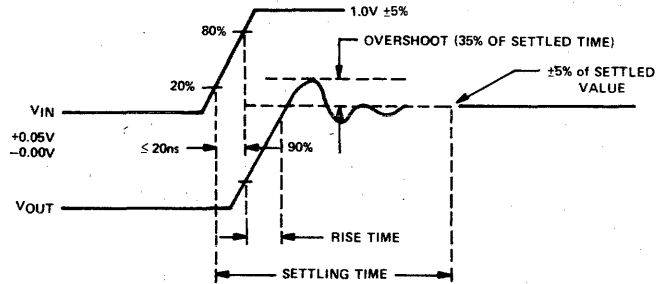
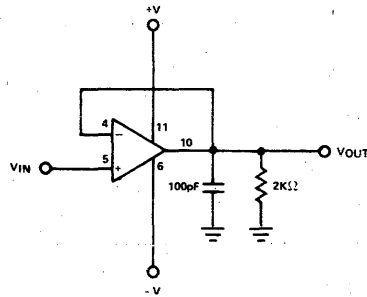
$$1. \text{ SLEW RATE} = \frac{5}{T_r^+ (\mu s)} \text{ V}/\mu s$$

$$\text{OR} - \frac{5}{T_r^- (\mu s)} \text{ V}/\mu s$$

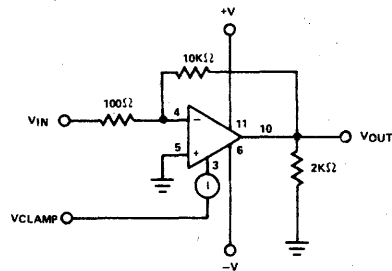
2. INPUT BIAS CURRENT OF NON-INVERTING INPUT MAY INCREASE IF  $V_{IN}$  APPLIED BEFORE +V AND -V.



### RISE TIME/SETTLING TIME CIRCUIT



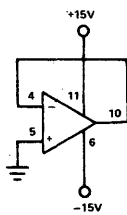
### VOLTAGE CLAMP CIRCUIT



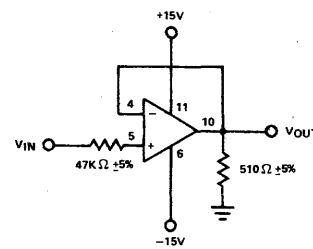
1.  $V_{IN} = +1.0 \text{ VDC}$
2.  $V_{CLAMP} = -3.0 \text{ VDC}$
3. MEASURE  $V_{OUT}$ ;  $V_{OUT}$  SHALL BE WITHIN  $\pm 0.4 \text{ VDC}$  OF  $V_{CLAMP}$ .
4. REPEAT STEPS 1, 2, 3 USING BOTH VOLTAGES OF OPPOSITE POLARITY.
5. REPEAT STEPS 1 THRU 4 USING A VALUE OF 6 VDC IN STEP 2.

	MIN	MAX
6. $I =$	-1.0mA	-3.3mA
	+0.5mA	+3.0mA

### IRRADIATION CIRCUIT



### BURN-IN CIRCUIT



$V_{IN} = 50 \text{ KHz SQUARE WAVE 50\% DUTY CYCLE -4.0V TO +4.0V}$

## Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^6$  Rad (Si)  $\pm 10\%$  from a gamma cell 220 cobalt 60 source or equivalent. The devices will be powered in the configuration illustrated with VSUPPLY =  $\pm 15V$ . The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4) AVOL, VIO, and IBIAS, with VSUPPLY =  $\pm 15V$ , will be measured and recorded for each device within one hour after irradiation. The lot will be accepted only if the sample, exclusive of non-radiation failures, meets the limits of AVOL  $\geq 80dB$ , VIO  $\leq \pm 5.0mV$  and IBIAS  $\leq \pm 400nA$  at room temperature.

## Radiation Effects

- (1) Total Dose  
Little or no effect will be observed at  $1 \times 10^5$  Rad (Si). IBIAS, IIO, and AVOL starts to degrade between  $1 \times 10^5$  and  $1 \times 10^6$  Rad (Si).
- (2) Dose Rate  
Devices are constructed in DI and consequently are latchup free.
- (3) Neutron Fluence  
Performance degradation is insignificant at  $5 \times 10^{12}n/cm^2$ .

## Test Product Flow

HARRIS SEMICONDUCTOR PRODUCT FLOW  
MIL-STD-883, METHOD 5004 CLASS B

100% SCREENING PROCEDURE

	SCREEN	MIL-STD-883 METHOD/COND.
①	Internal Visual	2010 Cond. B.
②	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
③	Temperature Cycling	1010 Cond. C
④	Constant Acceleration	2001 Cond. E; Y1 plane
⑤	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
⑥	Initial Electrical	Harris Specifications
⑦	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
⑧	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
⑨	External Visual	2009 Sample Inspection

Note:

- Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.
- Branding: All devices are branded with the part number and EIA date code.
- Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
- Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.

## Sales Offices

### EASTERN REGION

2600 Virginia Avenue  
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Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
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Minneapolis, MN 55420  
(612) 854-3224  
TWX: 910-576-3418

### HOME OFFICE

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174

CUSTOM INTEGRATED CIRCUITS DIVISION



**HARRIS**



# HARRIS

## HS-3530RH

### Low Power- Radiation Resistant Programmable Operational Amplifier

#### Features

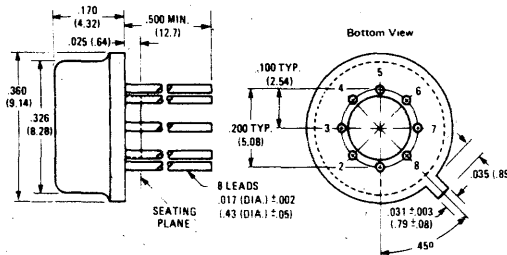
- WIDE RANGE AC PROGRAMMING  
SLEW RATE . . . . . .06 TO 3V/ $\mu$ s  
GAIN X BANDWIDTH . . . . . 100KHz TO 5.0MHz
- WIDE RANGE DC PROGRAMMING  
POWER SUPPLY RANGE . . . . .  $\pm$ 1.5V TO  $\pm$ 18V  
SUPPLY CURRENT . . . . . 10 $\mu$ A TO 1.2mA
- SHORT CIRCUIT PROTECTION
- RADIATION ENVIRONMENT  
NEUTRON FLUENCE( $\phi$ ) . . . . .  $5 \times 10^{12}$  n/cm<sup>2</sup> ( $E \geq 10$ KeV)  
GAMMA RATE( $\dot{\gamma}$ ) . . . . .  $1 \times 10^9$  RADS (Si)/s  
GAMMA DOSE( $\dot{\gamma}$ ) . . . . .  $1 \times 10^6$  RADS (Si)

#### Description

The HS-3530 is a Low Power Operational Amplifier which is an internally compensated monolithic device offering a wide range of performance specifications. Parameters such as power dissipation, slew rate, bandwidth, noise and input DC parameters are programmed by selecting an external resistor or current source. Supply voltages as low as  $\pm$ 3 volts may be used with little degradation of AC performance. The HS-3530 has been specifically designed to meet exposure to radiation environments. Operation from  $-55^\circ\text{C}$  to  $+125^\circ\text{C}$  is guaranteed.

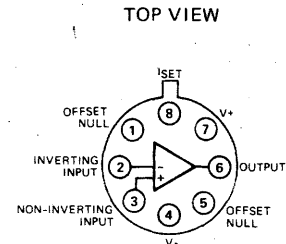
A major advantage of the HS-3530 is that operating characteristics remain virtually constant over a wide supply range (+3V to +15V), allowing the amplifier to offer maximum performance in almost any system, including battery operated equipment. A primary application for this device is in active filtering and conditioning for a wide variety of signals that differ in frequency and amplitude. Also, by modulating the set current, it can be used for designs such as current controlled oscillators/modulators, sample and hold circuits and variable active filters.

#### Package



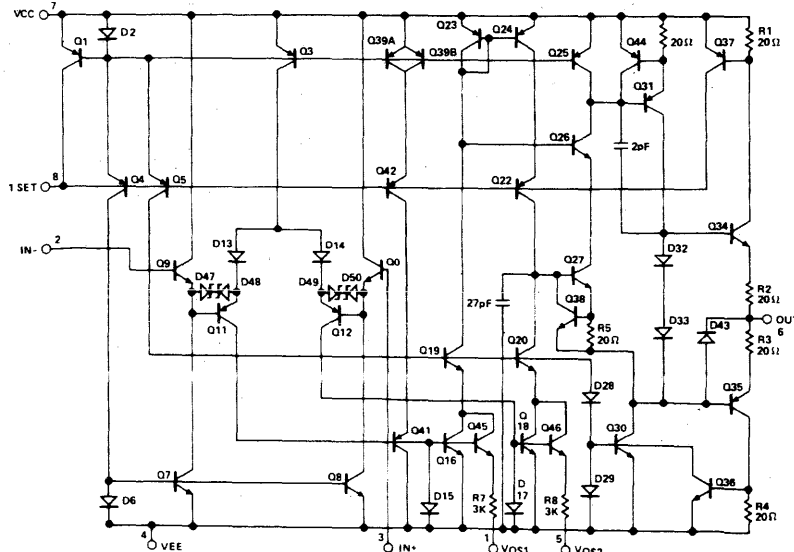
1. All dimensions in inches; millimeters are shown in parentheses.
2. All dimensions  $\pm .010$  ( $\pm 0.25$ mm) unless otherwise shown.

#### Pinout



NOTE: Case tied to V-

#### Schematic



# Specifications HS-3530

## ABSOLUTE MAXIMUM RATINGS

		ISET (current at ISET)	500 $\mu$ A
Voltage between V+ and V- terminals	40V	VSET (voltage to ground at ISET) (V+ -2.0V) < VSET < V+	
Differential Input Voltage	$\pm 20V$	Output Short Circuit Duration	Indefinite
Input Voltage (Note 1)	$\pm 15V$	Storage Temperature Range	-65°C to +150°C
		Operating Temperature Range	-55°C to +125°C

NOTE:

① For supply voltage less than  $\pm 15V$ , the absolute maximum input voltage is equal to the supply voltage.

## PRE-RADIATION ELECTRICAL CHARACTERISTICS

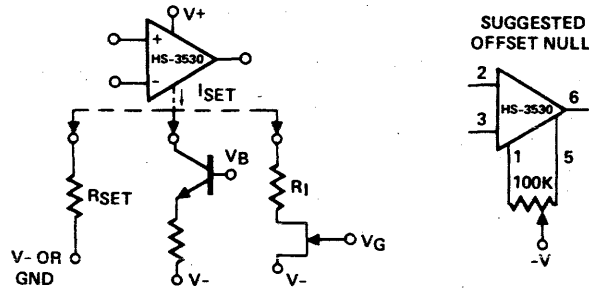
PARAMETER	TEMP.	ISET = 1.5 $\mu$ A (RL = 75K $\Omega$ )			ISET = 15 $\mu$ A (RL = 5K $\Omega$ )			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Offset Voltage	25°C		1	3		1	3	mV
	Full			5			5	mV
Input Bias Current	25°C		1.2			10	20	nA
	Full		2.5			22	40	nA
Input Offset Current	25°C		0.3			2	5	nA
	Full						10	nA
Input Resistance	25°C		>100		50	>100		M $\Omega$
Large Signal Voltage Gain <sup>1</sup>	25°C	65K	115K		80K	130K		V/V
	Full	25K	60K		50K	70K		V/V
Common Mode Rejection Ratio <sup>2</sup>	25°C	80	115		80	115		dB
	Full				80	110		dB
Supply Current	25°C		13	15		125	150	$\mu$ A
	Full		14	15		130	160	$\mu$ A
Power Supply Rejection Ratio <sup>3</sup>	25°C	80	130		80	130		dB
	Full				80	120		dB
Output Voltage Swing <sup>2</sup>	25°C	$\pm 12.5$	$\pm 14.2$		$\pm 12.5$	$\pm 14.2$		V
	Full	$\pm 10.5$	$\pm 14.0$		$\pm 10.5$	$\pm 14.0$		V
Output Current	25°C		$\pm 5$			$\pm 5$		mA
	Full		$\pm 4$			$\pm 4$		mA
Output Short-Circuit Current	25°C		2			14		mA
Gain-Bandwidth Product	25°C		85			850		kHz
	Full		65			640		kHz
Slew Rate <sup>5</sup>	25°C		.05			.55		V/ $\mu$ S
Rise Time <sup>4</sup>	25°C		7.5			.7		$\mu$ S
Overshoot <sup>4</sup>	25°C		5			10		%
Overdrive Recovery Time	25°C					2		$\mu$ S

PARAMETER	TEMP.	ISET = 1.5 $\mu$ A (RL = 75K $\Omega$ )			ISET = 15 $\mu$ A (RL = 5K $\Omega$ )			UNITS
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Large Signal Voltage Gain <sup>1</sup>	25°C	25K	60K		25K	75K		V/V
	Full	15K	40K		25K	50K		V/V
Common Mode Rejection Ratio <sup>2</sup>	25°C	80	100		80	95		dB
	Full	80	90					dB
Supply Current	25°C		12	15		115	150	$\mu$ A
	Full			15			160	$\mu$ A
Power Supply Rejection Ratio <sup>3</sup>	25°C	80	105		80	105		dB
	Full	80	100					dB
Output Voltage Swing <sup>1</sup>	25°C	$\pm 2.0$			$\pm 2.0$			V
	Full	$\pm 2.0$			$\pm 2.0$			V
Gain-Bandwidth Product	25°C		72			730		kHz
	Full		60			600		kHz
Slew Rate <sup>5</sup>	25°C		.04			.4		V/ $\mu$ S
Offset Voltage	25°C		1	3		1	3	mV
	Full			5			5	mV

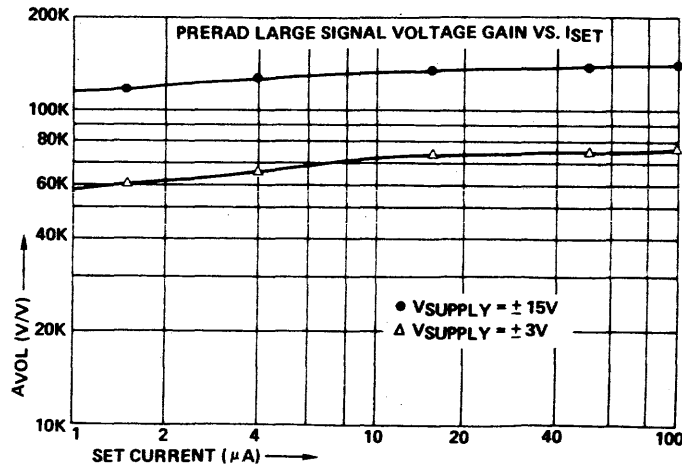
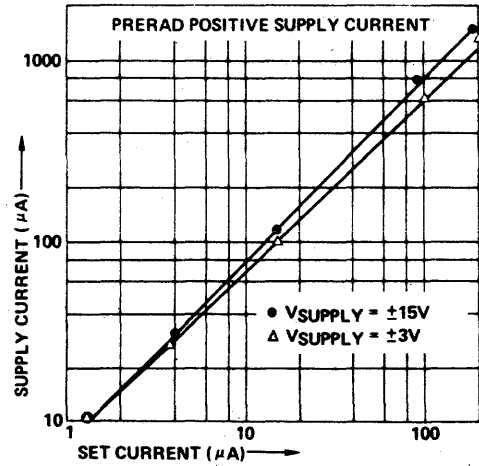
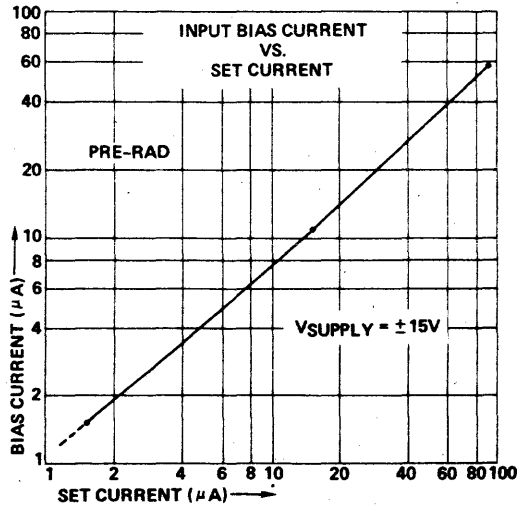
NOTES:

- |    |   |  |  |   |
|----|---|--|--|---|
| 1. | $\frac{V_{SUPPLY} = \pm 3.0V}{V_O = \pm 1.0V}$        | $\frac{V_{SUPPLY} = \pm 15.0V}{V_O = \pm 10.0V}$ | $\frac{I_{SET} = 1.5\mu A}{R_L = 75K}$ | $\frac{I_{SET} = 15\mu A}{R_L = 5K + 25^\circ C + 125^\circ C}$<br>$R_L = 75K - 55^\circ C$ |
| 2. | $V_{CM} = \pm 1.5V$                                   | $V_{CM} = \pm 5.0V$                              |  |   |
| 3. | $V = \pm 1.5V$  | $\Delta V = \pm 5.0V$                            |  |   |
| 4. | ← $A_V = +1, V_{IN} = 400mV, R_L = 5K, C_L = 100pF$ → |  |  |   |
| 5. | $V_O = \pm 2.0V$                                      | $V_O = \pm 10.0V$                                | $R_L = 20K$                            | $R_L = 5K + 25^\circ C + 125^\circ C$<br>$R_L = 75K - 55^\circ C$                           |

**Typical Biasing Circuits**

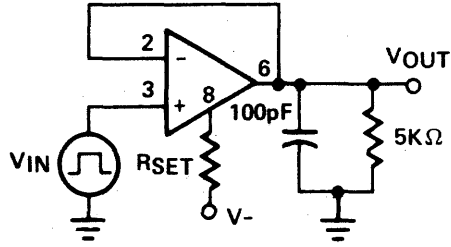


**Typical Performance Curves**

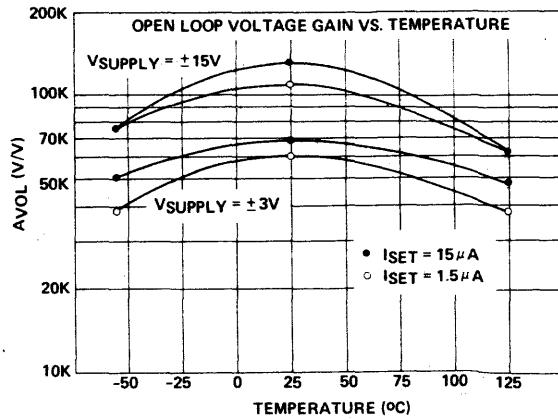
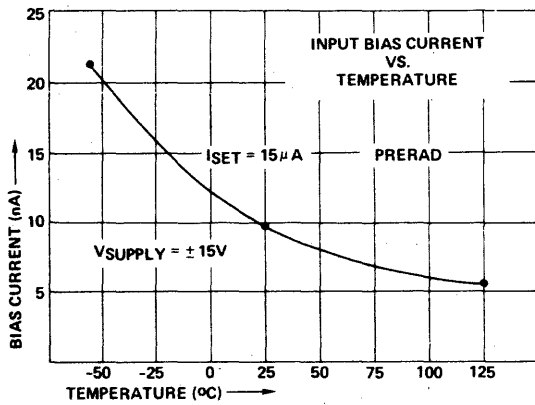
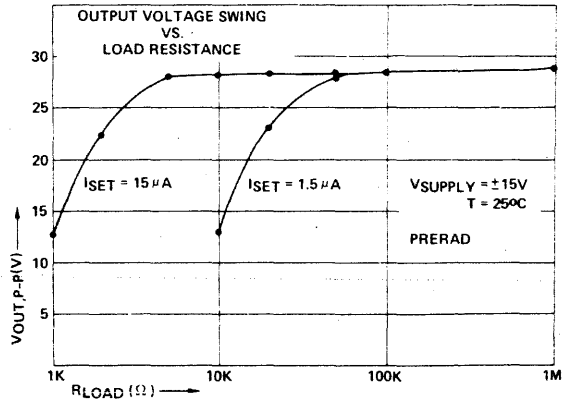
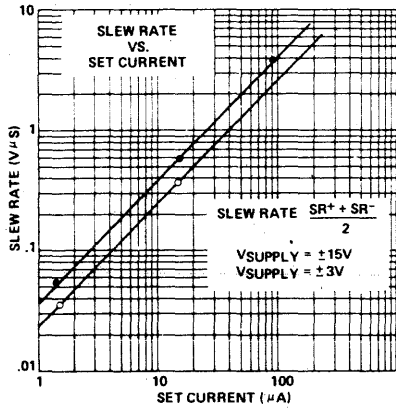
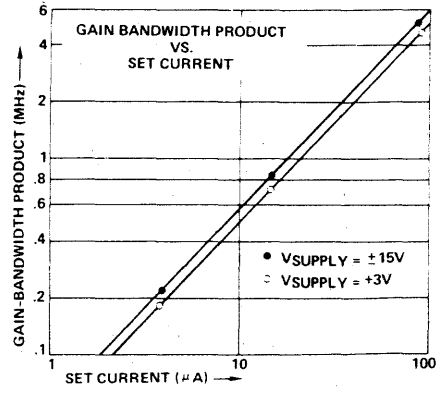




**Transient Response/Slew Rate Circuit**



**Typical Performance Curves**





### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^6$  Rad (Si)  $\pm 10\%$  from a Gamma Cell 220 Cobalt 60 source or equivalent. The devices will be powered in a voltage follower configuration, with  $I_{SET} = 15 \mu A$  and  $V_{SUPPLY} = \pm 15V$ . The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4)  $A_{VOL}$  and  $V_{IO}$ , with  $I_{SET} = 15 \mu A$ ,  $V_{SUPPLY} = +15V$  and  $R_L = 25K$ , will be measured and recorded for each device within one hour after irradiation. The lot will be accepted only if the sample, exclusive on non-radiation failures, meets the limits of  $A_{VOL} \geq 20K V/V$  and  $V_{IO} \leq 3.5 mV$  at room temperature.

### Radiation Effects

- (1) Total Dose: Very little degradation of any of the parameters will be seen up to  $\gamma = 10^4$  Rad (Si). Moderate degradation of open loop gain, bias current, and offset current begins at greater than  $10^5$  and less than  $10^6$  Rad (Si).
- (2) Dose Rate: During transient ionizing radiation at a level of  $1 \times 10^9$  rad (Si)/s, the peak level of supply current will be about 150 to 200 mA. After about 0.1 to 0.5  $\mu s$  this current drops about 10%. Maximum recovery time will be about 6 to 8  $\mu s$ .
- (3) Neutron Fluence: Large signal voltage gain degrades rapidly for low supply voltages and low set currents. For  $V_{SUPPLY} = +15V$  and  $I_{SET} > 10 \mu A$ , large signal voltage gain degrades by only about 50%. Input bias current doubles for  $\phi = 5 \times 10^{12} n/cm^2$ .

### Test Product Flow

**HARRIS SEMICONDUCTOR PRODUCT FLOW**  
**MIL - STD - 883, METHOD 5004 CLASS B**  
**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND.
①	Internal-Visual	2010 Cond. B.
②	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
③	Temperature Cycling	1010 Cond. C
④	Constant Acceleration	2001 Cond. E; Y1 plane
⑤	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
⑥	Initial Electrical	Harris Specifications
⑦	Burn-In Test	1015, 160 hrs. 125°C (or equivalent)
⑧	Final Electrical 100% go-no-go	Tested at Worst Case Operation Conditions
⑨	External Visual	2009 Sample Inspection

- Note: Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.
- Branding: All devices are branded with the part number and E1A date code.
- Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
- Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.



# HARRIS

## Preliminary

# HS-3546RH

## Radiation Resistant High Performance Operational Amplifier

### Features

- **LOW OFFSET VOLTAGE** 0.3mV
- **HIGH SLEW RATE**  $\pm 4V/\mu s$
- **WIDE BANDWIDTH** 8MHz
- **LOW DRIFT**  $2\mu V/^\circ C$
- **FAST SETTling (0.01%, 10V STEP)** 4.2 $\mu s$
- **LOW POWER CONSUMPTION** 35mW
- **SUPPLY RANGE**  $\pm 5V$  TO  $\pm 20V$
- **RADIATION ENVIRONMENT**
  - NEUTRON FLUENCE ( $\phi$ )  $.5 \times 10^{12}$  n/cm<sup>2</sup> (E  $\geq$  10KeV)
  - GAMMA RATE ( $\dot{\gamma}$ )  $1 \times 10^9$  RADS (Si)/s
  - GAMMA DOSE ( $\gamma$ )  $1 \times 10^6$  RADS (Si)

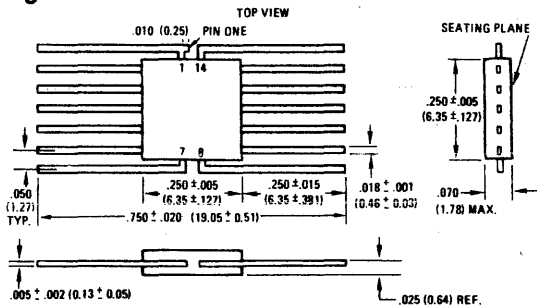
### Description

The HS-3546RH is a radiation resistant, high performance dielectrically isolated monolithic operational amplifier with superior specifications. This amplifier offers excellent dynamic performance coupled with low values for offset voltage and drift, input noise voltage and power consumption.

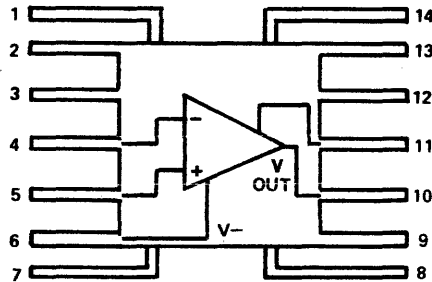
A wide range of applications can be achieved by using the features made available by the HS-3546RH. With wide bandwidth (8MHz), low power (35mW) and internal compensation, these devices are ideally suited for precision active filter designs. For audio applications these amplifiers offer low noise (8nV/ $\sqrt{Hz}$ ) and excellent full power bandwidth (60 KHz). The HS-3546RH is particularly useful in designs requiring low offset voltage (0.3mV) and drift (2 $\mu V/^\circ C$ ), such as instrumentation and signal conditioning circuits. The high slew rate (4V/ $\mu s$ ) and fast settling time (4.2  $\mu s$  to 0.01%, 10V step) makes this amplifier a useful component in fast, accurate data acquisition systems.

The HS-3546RH has been specifically designed to meet exposure to radiation environments. It is available in a 14 Pin Ceramic Flat-pack package and is guaranteed operational from -55 $^\circ C$  to +125 $^\circ C$ .

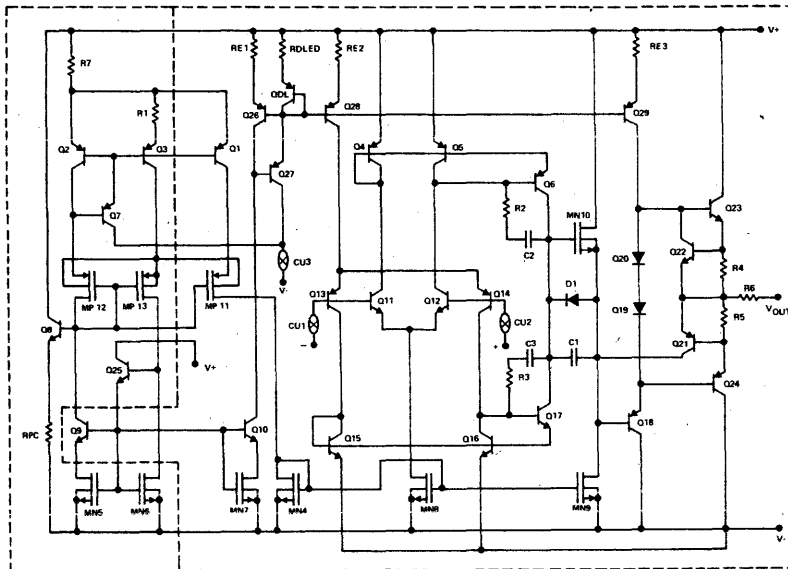
### Package



### Pinout



### Schematic



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**ABSOLUTE MAXIMUM RATINGS (Note 1)**

$T_A = +25^\circ\text{C}$ Unless Otherwise Stated		Power Dissipation (Note 4)	880mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	$-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$
Differential Input Voltage	$\pm 7\text{V}$	Storage Temperature Range	$-65^\circ\text{C} \leq T_A \leq +150^\circ\text{C}$
Input Voltage (Note 2)	$\pm 15.0\text{V}$		
Output Short Circuit Duration (Note 3)	Indefinite		

**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

**ELECTRICAL CHARACTERISTICS**  $V_+ = 15\text{V}$ ,  $V_- = -15\text{V}$   $T_A = -55^\circ\text{C}$  to  $+125^\circ\text{C}$

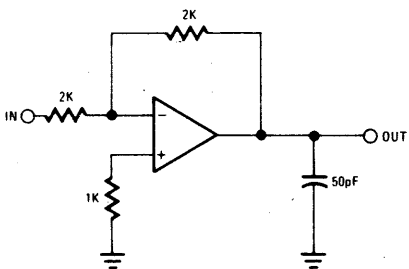
PARAMETER	TEMP	MIN	TYP	MAX	UNITS
<b>INPUT CHARACTERISTICS</b>					
* Offset Voltage	+25°C		0.3	2.5	mV
	Full			3.0	mV
Av. Offset Voltage Drift	Full		2		$\mu\text{V}/^\circ\text{C}$
* Bias Current	+25°C		130	200	nA
	Full			325	nA
* Offset Current	+25°C		30	75	nA
	Full			125	nA
Common Mode Range	Full	$\pm 12$			V
Input Noise Voltage (f = 1kHz)	+25°C		8		$\text{nV}/\sqrt{\text{Hz}}$
Input Resistance			500		$\text{k}\Omega$
<b>TRANSFER CHARACTERISTICS</b>					
* Large Signal Voltage Gain (Note 5)	Full	50K	150K		V/V
* Common Mode Rejection Ratio (Note 8)	Full	86			dB
Small Signal Bandwidth	+25°C		8		MHz
<b>OUTPUT CHARACTERISTICS</b>					
* Output Voltage Swing ( $R_L = 10\text{K}$ )	Full	$\pm 12$	$\pm 13$		V
( $R_L = 2\text{K}$ )	Full	$\pm 10$	$\pm 12$		V
Full Power Bandwidth (Note 5)	+25°C		60		kHz
Output Current (Note 6)	Full	$\pm 18$	$\pm 25$		mA
Output Resistance	+25°C		200		$\Omega$
<b>TRANSIENT RESPONSE (Note 7)</b>					
Rise Time	+25°C		50	150	ns
Overshoot	+25°C		30	45	%
Slew Rate	+25°C	$\pm 1$	$\pm 4$		$\text{V}/\mu\text{s}$
Settling Time (Note 9)			4.2		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>					
* Supply Current	+25°C		4.6	5.5	mA
* Power Supply Rejection Ratio (Note 8)	Full	86			dB

\*100% tested

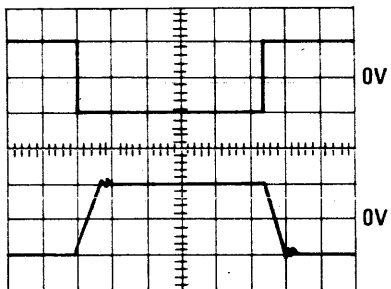
**NOTES:**

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
2. For supply voltages less than  $\pm 15V$ , the absolute maximum input voltage is equal to the supply voltage.
3. Any one amplifier may be shorted to ground indefinitely.
4. Derate  $5.8mW/^{\circ}C$  above  $T_A = +25^{\circ}C$ .
5.  $V_{OUT} = \pm 10V$ ;  $R_L = 2K$  ohms.
6. Output current is measured with  $V_{OUT} = \pm 5$  volts.
7. For transient response test circuits and measurement conditions refer to Test Circuits section of the data sheet.
8.  $\Delta V = \pm 5.0$  volts.
9. Settling time is measured to 0.1% of final value for a 10 volt input step,  $A_V = -1$ .

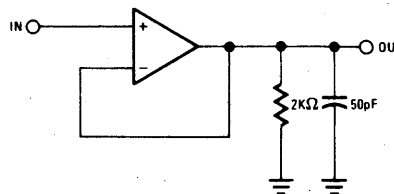
**Test Circuits**



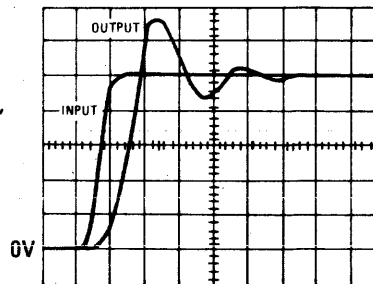
**LARGE SIGNAL RESPONSE CIRCUIT**  
(Volts: 5V/Div., Time: 5 $\mu$ s/Div.)



VERT. 5V/DIV.  
HORZ. 5 $\mu$ s/DIV.

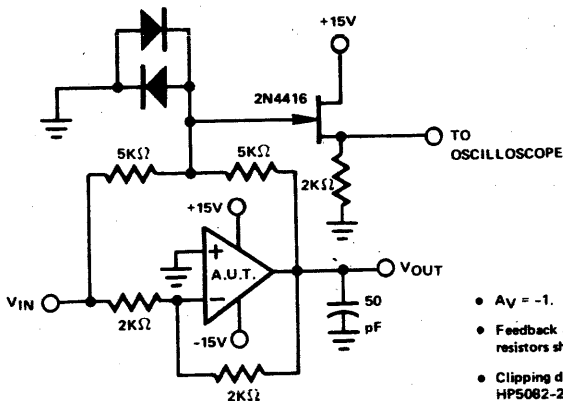


**SMALL SIGNAL RESPONSE CIRCUIT**  
(Volts: 10mV/Div., Time: 50ns/Div.)



HORIZONTAL: 50 NSEC/DIV.  
VERTICAL: 10mV/DIV

**SETTLING TIME CIRCUIT**



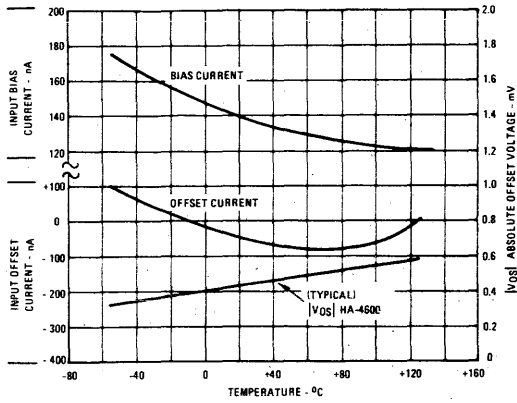
- $A_V = -1$ .
- Feedback and summing resistors should be 0.1%.
- Clipping diodes are optional. HP5082-2810 recommended.

# Performance Curves

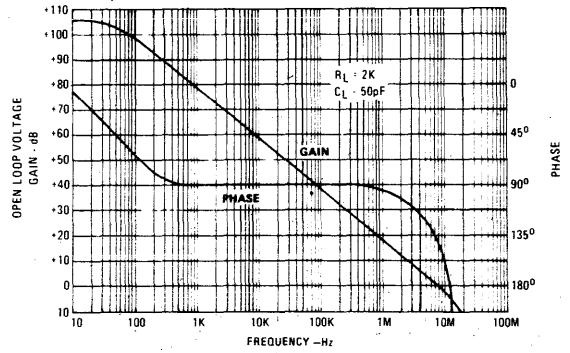


$V_+ = +15V$ ,  $V_- = -15V$ ,  $T_A = +25^\circ C$  Unless Otherwise Stated.

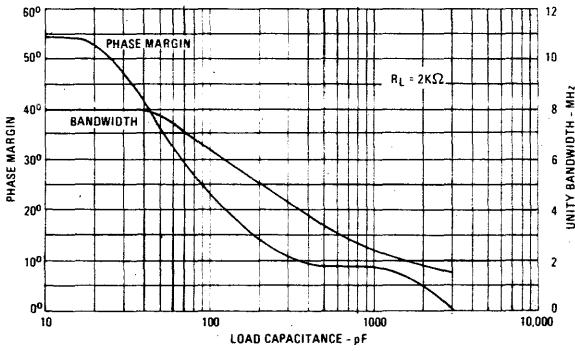
### OFFSET VOLTAGE INPUT BIAS AND OFFSET CURRENT VS. TEMPERATURE



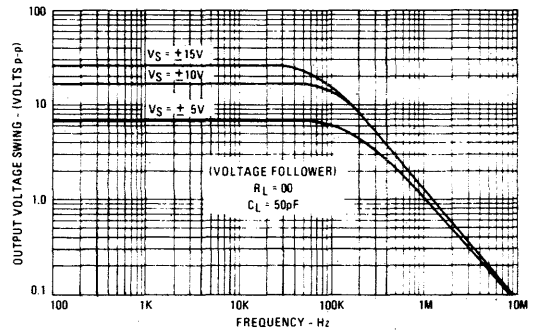
### OPEN LOOP FREQUENCY RESPONSE



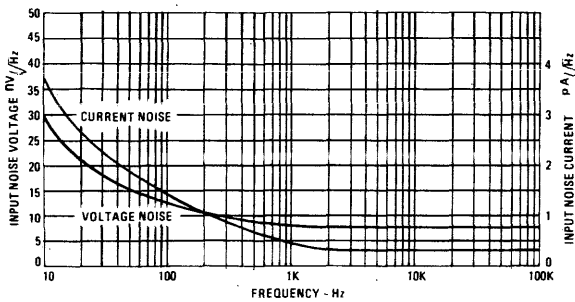
### SMALL SIGNAL BANDWIDTH AND PHASE MARGIN VS. LOAD CAPACITANCE



### OUTPUT VOLTAGE SWING VS. FREQUENCY AND SUPPLY VOLTAGE



### INPUT NOISE VS. FREQUENCY

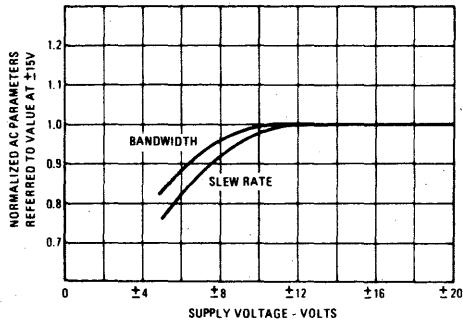


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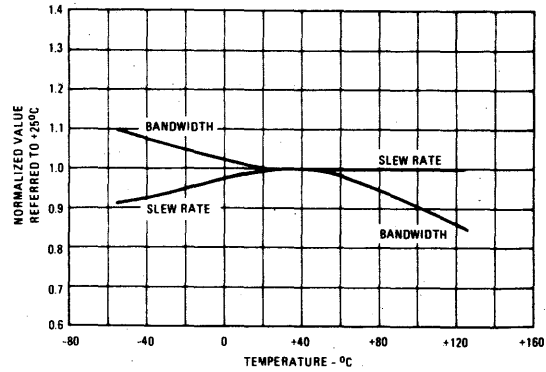
CUSTOM/SEMICUSTOM



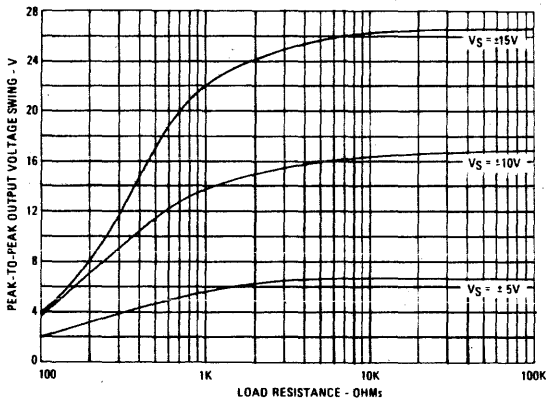
NORMALIZED AC PARAMETERS VS. SUPPLY VOLTAGE



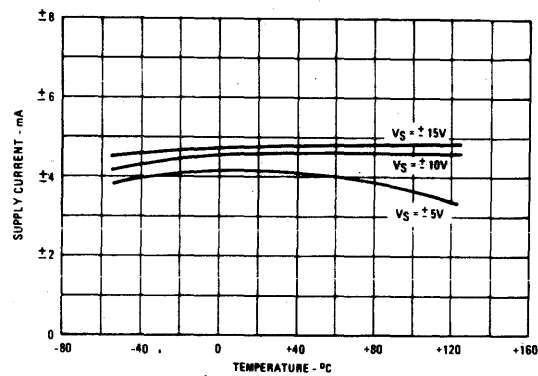
NORMALIZED AC PARAMETERS VS. TEMPERATURE



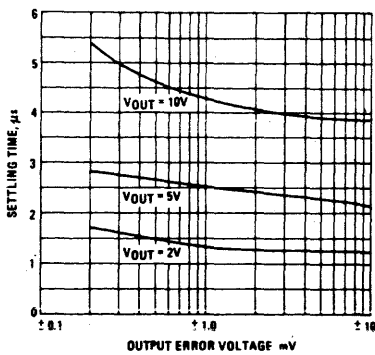
MAXIMUM OUTPUT VOLTAGE SWING VS. LOAD RESISTANCE AND SUPPLY VOLTAGE



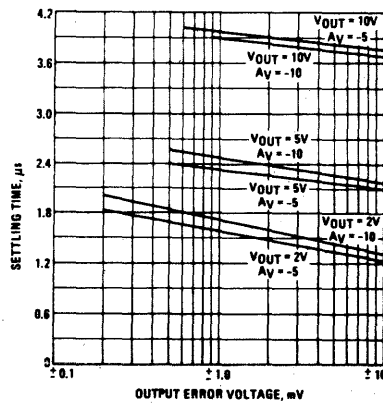
POWER SUPPLY CURRENT VS. TEMPERATURE AND SUPPLY VOLTAGE



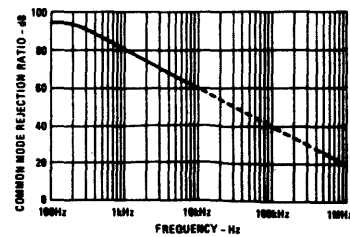
SETTLING TIME VS. OUTPUT AMPLITUDE (A\_V = -1)



SETTLING TIME VS. OUTPUT AMPLITUDE AND SIGNAL GAIN (A\_V = -5 AND A\_V = -10)



COMMON MODE REJECTION RATIO VS. FREQUENCY



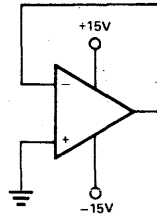
Harris Semiconductor

CUSTOM/SEMICUSTOM

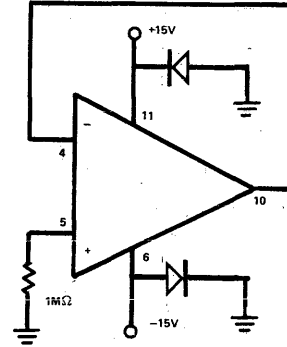
1. **POWER SUPPLY DECOUPLING:** Although not absolutely necessary, it is recommended that all power supply lines be decoupled with .01  $\mu$ F ceramic capacitors to ground. Decoupling capacitors should be located as near to the amplifier terminals as possible.

2. In high frequency applications where large value feedback resistors are used, a small capacitor (3pF) may be needed in parallel with the feedback resistor to neutralize the pole introduced by input capacitance.

### Irradiation Circuit



### Burn-in Circuit



NOTES:  
 $T_A = +125^{\circ}\text{C}$   
 D = IN4002 OR EQUIVALENT

### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^5$  Rad (Si)  $\pm 10\%$  from a gamma cell 220 cobalt 60 source or equivalent. The devices will be powered in the configuration illustrated with  $V_{\text{SUPPLY}} = \pm 15\text{V}$ . The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4)  $AVOL$ ,  $V_{\text{IO}}$ , AND  $I_{\text{BIAS}}$ , with  $V_{\text{SUPPLY}} = \pm 15\text{V}$ , will be measured and recorded for each device within one hour after irradiation. The lot will be accepted only if the sample, exclusive of non-radiation failures, meets the limits of  $AVOL \geq 50\text{K}$ ,  $V_{\text{IO}} \leq \pm 2.5\text{mV}$  and  $I_{\text{BIAS}} \leq 1.0 \mu\text{A}$  at room temperature.

### Radiation Effects

- (1) Total Dose  
 Little or no effect will be observed at  $1 \times 10^4$  Rad (Si).  $I_{\text{BIAS}}$ ,  $I_{\text{IO}}$  and  $AVOL$  starts to degrade between  $1 \times 10^4$  and  $1 \times 10^5$  Rad (Si).
- (2) Dose Rate  
 Devices are constructed in DI and consequently are latchup free.



**Test Product Flow**

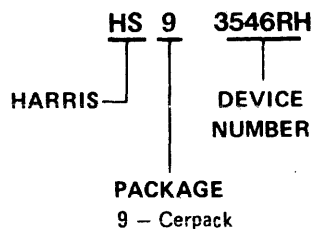
**HARRIS SEMICONDUCTOR PRODUCT FLOW  
MIL-STD-883, METHOD 5004 CLASS B**

**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

- Notes:**
- Traceability:** All devices are assigned date code identification that provides traceability back to the inspection lot.
  - Branding:** All devices are branded with the part number and EIA date code.
  - Aged Products:** Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
  - Additional Requirements:** Sample Group A electrical tests are performed on a lot acceptance basis.

**Ordering Information**



**Sales Offices**

**UNITED STATES**

**EASTERN REGION**

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
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Five Old Concord Road  
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(617) 273-1020  
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(516) 747-6776

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TWX: 910-576-3418

**HOME OFFICE**

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

**EUROPEAN**

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



**HARRIS**

**CUSTOM INTEGRATED CIRCUITS DIVISION**



**HARRIS**

*Preliminary*

# HS-4602RH

Radiation Resistant  
High Performance  
Quad Operational Amplifier

## Features

- LOW OFFSET VOLTAGE 0.3mV
- HIGH SLEW RATE  $\pm 4V/\mu s$
- WIDE BANDWIDTH 8MHz
- LOW DRIFT  $2\mu V/^\circ C$
- FAST SETTLING (0.01%, 10V STEP) 4.2 $\mu s$
- LOW POWER CONSUMPTION 35mW/AMP
- SUPPLY RANGE  $\pm 5V$  TO  $\pm 20V$
- RADIATION ENVIRONMENT
  - NEUTRON FLUENCE ( $\phi$ ) . . . . .  $5 \times 10^{12}$  n/cm<sup>2</sup> (E  $\geq$  10KeV)
  - GAMMA RATE ( $\dot{\gamma}$ ) . . . . .  $1 \times 10^9$  RADS (Si)/s
  - GAMMA DOSE ( $\gamma$ ) . . . . .  $1 \times 10^6$  RADS (Si)

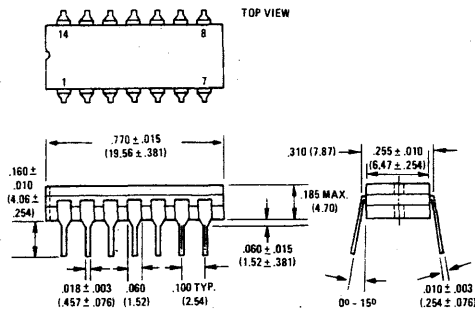
## Description

The HS-4602RH is a radiation resistant, high performance dielectrically isolated monolithic quad operational amplifier with superior specifications not previously available in a quad amplifier. This amplifier offers excellent dynamic performance coupled with low values for offset voltage and drift, input noise voltage and power consumption.

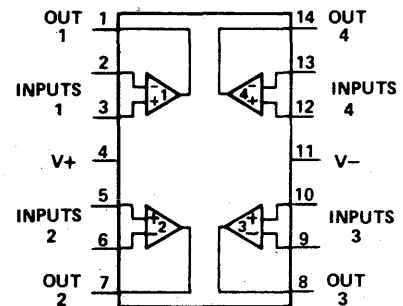
A wide range of applications can be achieved by using the features made available by the HS-4602RH. With wide bandwidth (8MHz), low power (35mW/amp), and internal compensation, these devices are ideally suited for precision active filter designs. For audio applications these amplifiers offer low noise ( $8nV/\sqrt{Hz}$ ) and excellent full power band-width (60KHz). The HS-4602RH is particularly useful in designs requiring low offset voltage (0.3mV) and drift ( $2\mu V/^\circ C$ ), such as instrumentation and signal conditioning circuits. The high slew rate ( $4V/\mu s$ ) and fast settling time (4.2 $\mu s$  to 0.01%, 10V step) makes this amplifier a useful component in fast, accurate data acquisition systems.

The HS-4602RH has been specifically designed to meet exposure to radiation environments. It is available in a 14 pin dual-in-line package and is guaranteed operational from  $-55^\circ C$  to  $+125^\circ C$ .

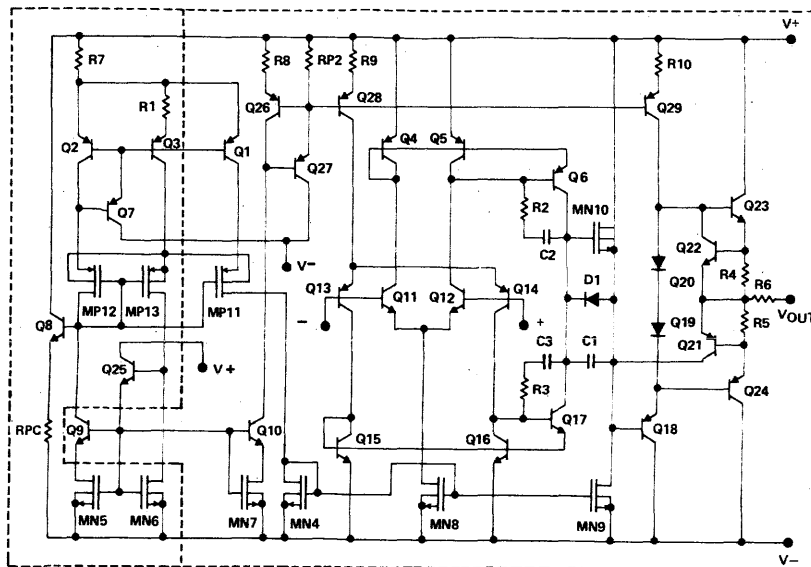
## Package



## Pinout



## Schematic



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(ONE FOURTH ONLY HS-4602-RH)

Harris Semiconductor

CUSTOM/SEMICUSTOM



# Specifications HS-4602RH

## ABSOLUTE MAXIMUM RATINGS (Note 1)

$T_A = +25^{\circ}\text{C}$ Unless Otherwise Stated		Power Dissipation (Note 4)	880mW
Voltage Between V+ and V- Terminals	40.0V	Operating Temperature Range	$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
Differential Input Voltage	$\pm 7\text{V}$	Storage Temperature Range	$-65^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$
Input Voltage (Note 2)	$\pm 15.0\text{V}$		
Output Short Circuit Duration (Note 3)	Indefinite		

**CAUTION:** To prevent permanent damage to this device, care should be exercised to insure that the absolute maximum ratings for supply voltages, temperature and voltage or current at any pin is not exceeded during both static and dynamic operation.

## ELECTRICAL CHARACTERISTICS $V_+ = 15\text{V}, V_- = -15\text{V}$ $T_A = -55^{\circ}\text{C}$ to $+125^{\circ}\text{C}$

PARAMETER	TEMP	MIN	TYP	MAX	UNITS
<b>INPUT CHARACTERISTICS</b>					
* Offset Voltage	+25°C		0.3	2.5	mV
	Full			3.0	mV
Av. Offset Voltage Drift	Full		2		$\mu\text{V}/^{\circ}\text{C}$
* Bias Current	+25°C		130	200	nA
	Full			325	nA
* Offset Current	+25°C		30	75	nA
	Full			125	nA
Common Mode Range	Full	$\pm 12$			V
Input Noise Voltage (f = 1kHz)	+25°C		8		$\text{nV}/\sqrt{\text{Hz}}$
Input Resistance			500		k $\Omega$
<b>TRANSFER CHARACTERISTICS</b>					
* Large Signal Voltage Gain (Note 5)	Full	50K	150K		V/V
* Common Mode Rejection Ratio (Note 9)	Full	86			dB
Channel Separation (Note 6)	+25°C		-108		dB
Small Signal Bandwidth	+25°C		8		MHz
<b>OUTPUT CHARACTERISTICS</b>					
* Output Voltage Swing ( $R_L = 10\text{K}$ )	Full	$\pm 12$	$\pm 13$		V
( $R_L = 2\text{K}$ )	Full	$\pm 10$	$\pm 12$		V
Full Power Bandwidth (Note 5)	+25°C		60		kHz
Output Current (Note 7)	Full	$\pm 10$	$\pm 15$		mA
Output Resistance	+25°C		200		$\Omega$
<b>TRANSIENT RESPONSE (Note 8)</b>					
Rise Time	+25°C		50	150	ns
Overshoot	+25°C		30	45	%
Slew Rate	+25°C	$\pm 1$	$\pm 4$		V/ $\mu\text{s}$
Settling Time (Note 10)			4.2		$\mu\text{s}$
<b>POWER SUPPLY CHARACTERISTICS</b>					
* Supply Current	+25°C		4.6	5.5	mA
* Power Supply Rejection Ratio (Note 9)	Full	86			dB

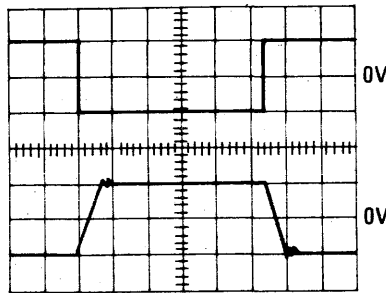
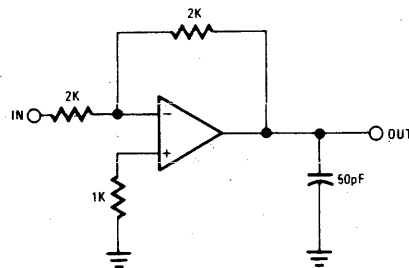
\*100% tested

**NOTES:**

1. Absolute maximum ratings are limiting values, applied individually, beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied.
2. For supply voltages less than  $\pm 15V$ , the absolute maximum input voltage is equal to the supply voltage.
3. Any one amplifier may be shorted to ground indefinitely.
4. Derate 5.8mW/°C above  $T_A = +25^\circ C$ .
5.  $V_{OUT} = \pm 10V$ ;  $R_L = 2K$  ohms.
6. Channel separation value is referred to the input of the amplifier. Input test conditions are:  $f = 10kHz$ ;  $V_{IN} = 200mV$  peak-to-peak;  $R_S = 1K$  ohms. (Refer to Channel Separation vs. Frequency Curve for test circuits.)
7. Output current is measured with  $V_{OUT} = \pm 5$  volts.
8. For transient response test circuits and measurement conditions refer to Test Circuits section of the data sheet.
9.  $\Delta V = \pm 5.0$  volts.
10. Settling time is measured to 0.1% of final value for a 10 volt input step,  $A_V = -1$ .

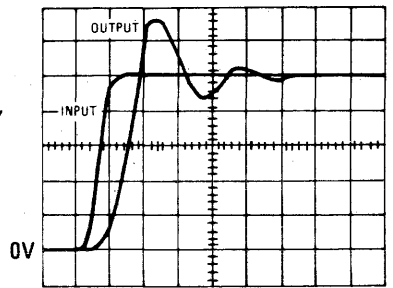
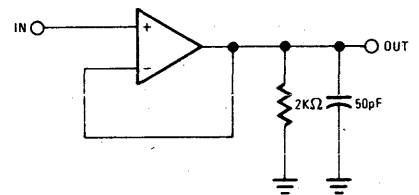
**Test Circuits**

**LARGE SIGNAL RESPONSE CIRCUIT**  
(Volts: 5V/Div., Time: 5 $\mu$ s/Div.)



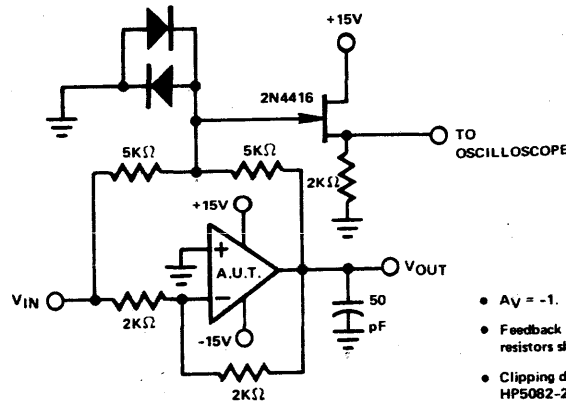
VERT. 5V/DIV.  
HORZ. 5  $\mu$ s/DIV.

**SMALL SIGNAL RESPONSE CIRCUIT**  
(Volts: 10mV/Div., Time: 50ns/Div.)



HORIZONTAL: 50 NSEC/DIV.  
VERTICAL: 10mV/DIV

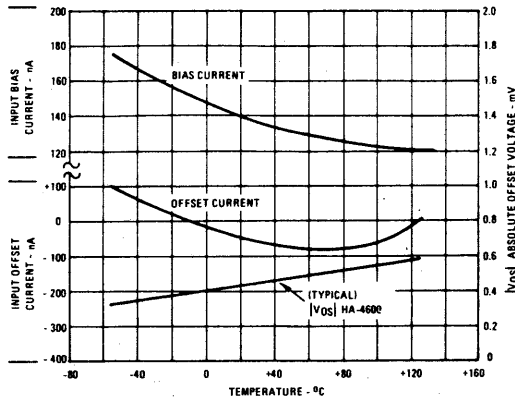
**SETTLING TIME CIRCUIT**



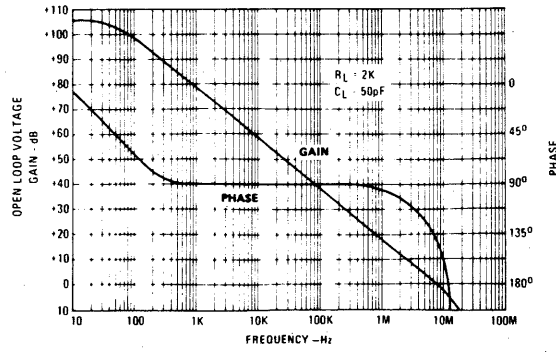
- $A_V = -1$ .
- Feedback and summing resistors should be 0.1%.
- Clipping diodes are optional. HP5082-2810 recommended.

V+ = +15V, V- = -15V, TA = +25°C Unless Otherwise Stated.

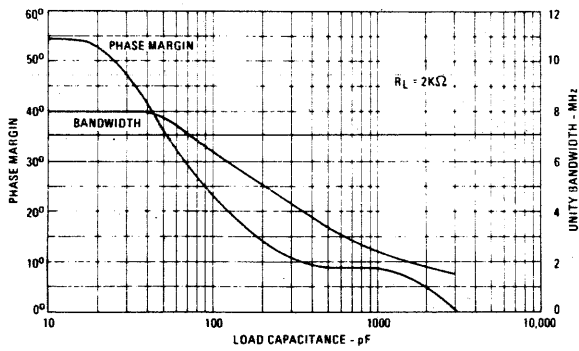
OFFSET VOLTAGE INPUT BIAS AND OFFSET CURRENT VS. TEMPERATURE



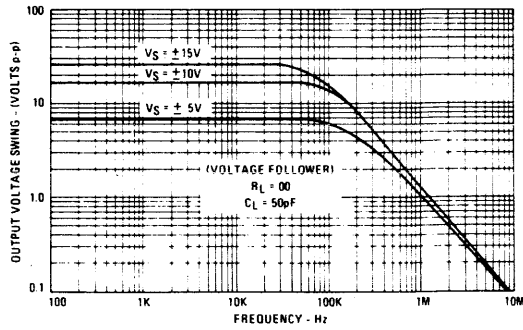
OPEN LOOP FREQUENCY RESPONSE



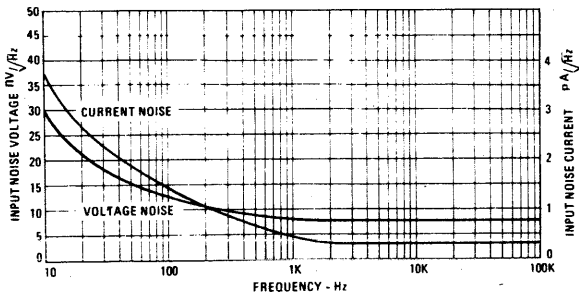
SMALL SIGNAL BANDWIDTH AND PHASE MARGIN VS. LOAD CAPACITANCE



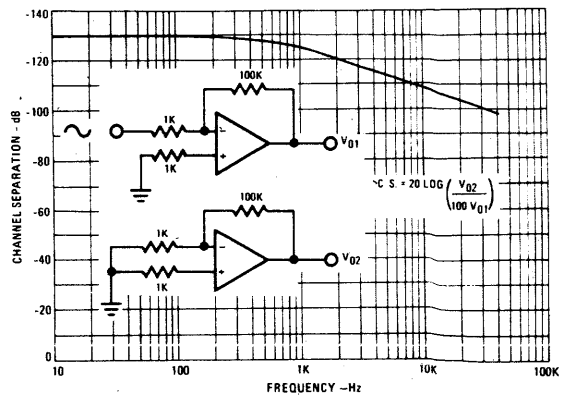
OUTPUT VOLTAGE SWING VS. FREQUENCY AND SUPPLY VOLTAGE



INPUT NOISE VS. FREQUENCY

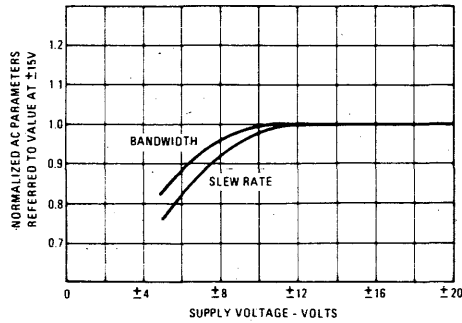


CHANNEL SEPARATION VS. FREQUENCY

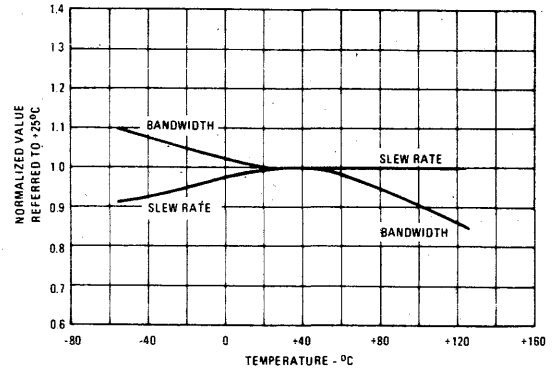




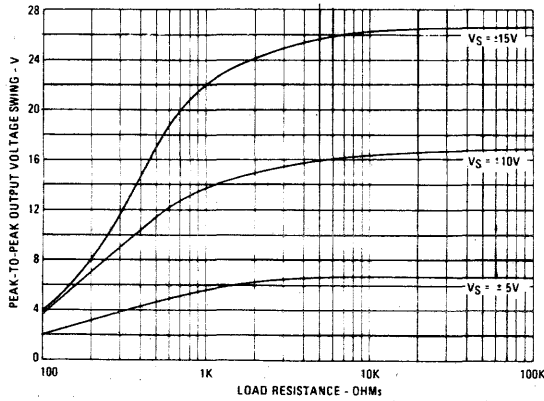
NORMALIZED AC PARAMETERS VS. SUPPLY VOLTAGE



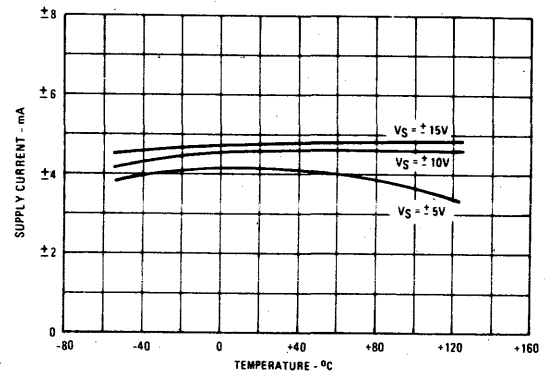
NORMALIZED AC PARAMETERS VS. TEMPERATURE



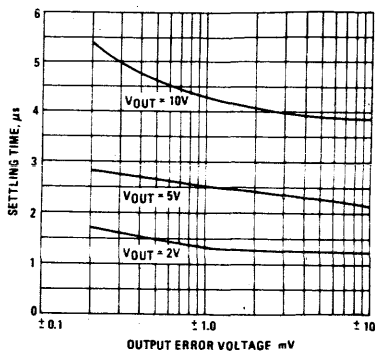
MAXIMUM OUTPUT VOLTAGE SWING VS. LOAD RESISTANCE AND SUPPLY VOLTAGE



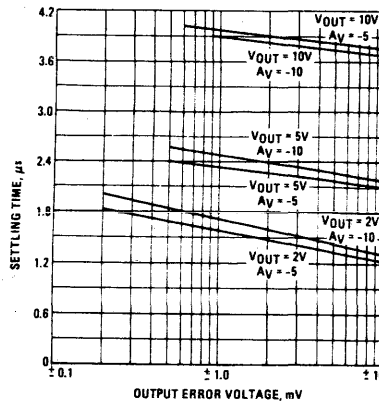
POWER SUPPLY CURRENT VS. TEMPERATURE AND SUPPLY VOLTAGE



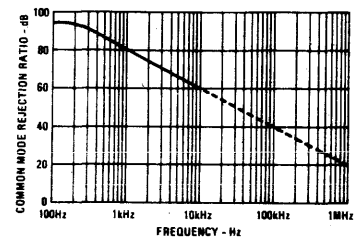
SETTLING TIME VS. OUTPUT AMPLITUDE ( $A_V = -1$ )



SETTLING TIME VS. OUTPUT AMPLITUDE AND SIGNAL GAIN ( $A_V = -5$  AND  $A_V = -10$ )



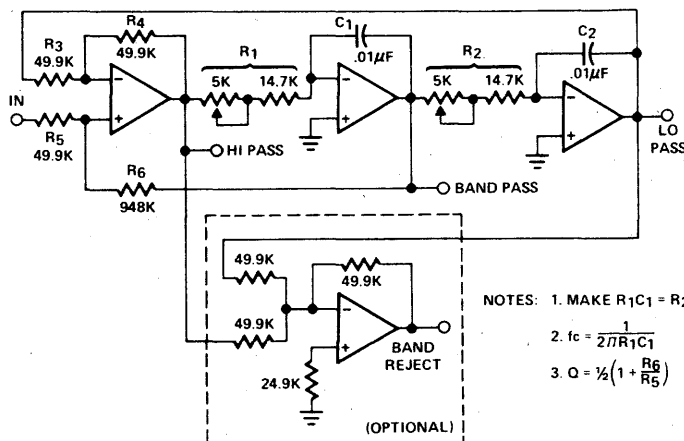
COMMON MODE REJECTION RATIO VS. FREQUENCY



1. **POWER SUPPLY DECOUPLING:** Although not absolutely necessary, it is recommended that all power supply lines be decoupled with .01  $\mu$  F ceramic capacitors to ground. Decoupling capacitors should be located as near to the amplifier terminals as possible.
2. **UNUSED OP AMPS:** Unused op amp sections should be connected in a non-inverting follower configuration with the (+) input tied to ground in order to insure optimum performance of devices being used.
3. In high frequency applications where large value feedback resistors are used, a small capacitor (3pF) may be needed in parallel with the feedback resistor to neutralize the pole introduced by input capacitance.

## Applications

### 2ND ORDER STATE VARIABLE FILTER (1kHz, Q = 10)



- NOTES:
1. MAKE  $R_1 C_1 = R_2 C_2$
  2.  $f_c = \frac{1}{2\pi R_1 C_1}$
  3.  $Q = \frac{1}{2} \left( 1 + \frac{R_6}{R_5} \right)$

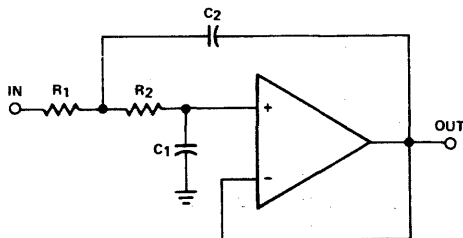
The state variable filter is relatively insensitive to component changes (changes can be adjusted out with potentiometers) and also has low sensitivity to amplifier bandwidths. (Amplifier gain bandwidth product should be  $\gg Q \times f_c$ ).

loose tolerances to be used. To tune for  $f_c$ , apply a sine wave at  $f_c$  to the input, adjust  $R_1$  for equal amplitudes at the Hi pass and Band pass terminals (they will be phased  $90^\circ$  apart) then adjust  $R_2$  for equal amplitudes at the Band pass and Lo pass terminals.

This filter finds wide application because multiple filtering functions are available simultaneously (High pass, Lo pass, Band pass, Band reject). In this circuit the various RC products are matched with pot adjustments allowing for non-interactive adjustment of Q and  $f_c$ . This allows capacitors ( $C_1, C_2$ ) with

The state variable filter is often used as building blocks in multiple pole Butterworth or Chebyshev filters. Many references contain normalized tables indicating settings for Q and  $f_c$  of each pole-pair section.

### SALLEN AND KEY 2ND ORDER LO PASS FILTER



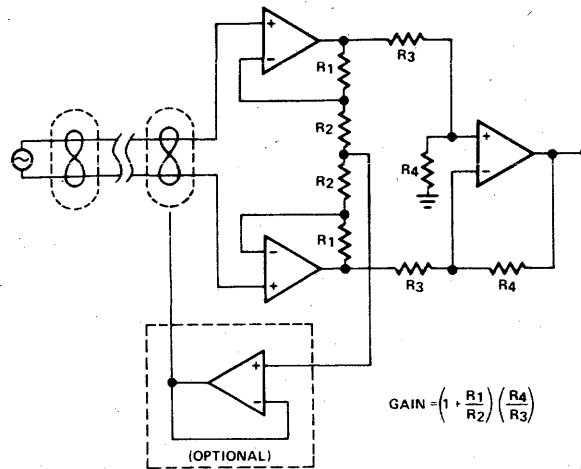
- NOTES:
1. Make  $R_1 = R_2$
  2.  $f_c = \frac{1}{2\pi R_1 \sqrt{C_1 C_2}}$
  3.  $Q = \frac{1}{2} \sqrt{\frac{C_2}{C_1}}$

The advantage of using the Sallen and Key filter is simplicity, but in any application this must be weighed against the state-variable type filter for accuracy, practicality, and cost. Amplifier bandwidth limitations are much more apparent at moderate frequencies and Q values with this filter design. (For accuracy, amplifier gain-bandwidth product should be  $\gg f_c \times Q^2$ ). The wide bandwidth of the HS-4602RH is particularly advantageous in this design even at audio frequencies.

In this filter all component values affect both Q and  $f_c$ . Precision, temperature stable resistors and capacitors must be used.

For economy, this filter could be used in the low Q stages of multiple-pole filter design, while the state variable type is used in the more critical stages.

### INSTRUMENTATION AMPLIFIER



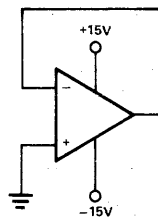
Instrumentation amplifiers (differential amplifiers) are specifically designed to extract and amplify small differential signals from much larger common mode voltages.

To serve as building blocks in instrumentation amplifiers, op amps must have very low offset voltage drift, high gain and wide bandwidth. The HS-4602RH is ideally suited for this appli-

cation, delivering superior input and speed characteristics.

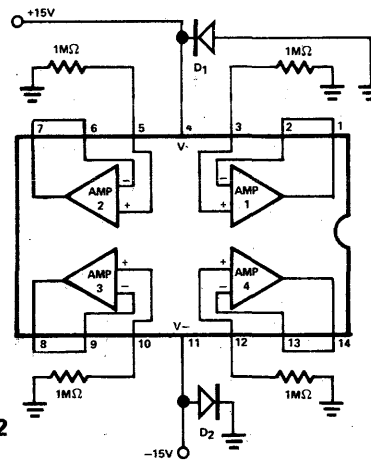
The optional circuitry makes use of the fourth amplifier section as a shield driver which enhances the AC common mode rejection by nullifying the effects of capacitance-to-ground mismatch between input conductors.

### Irradiation Circuit



(1/4 OF HS-4602RH DEPICTED)

### Burn-in Circuit



NOTES:  
 $T_A = +125^\circ\text{C}$   
 $D_1, D_2 = \text{IN4002}$

### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a run (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^5$  Rad (Si)  $\pm 10\%$  from a gamma cell 220 cobalt 60 source or equivalent. The devices will be powered in the configuration illustrated with  $V_{\text{SUPPLY}} = \pm 15\text{V}$ . The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4)  $AVOL$ ,  $V_{IO}$ , AND  $IBIAS$ , with  $V_{\text{SUPPLY}} = \pm 15\text{V}$ , will be measured and recorded for each device within one hour after irradiation. The lot will be accepted only if the sample, exclusive of non-radiation failures, meets the limits of  $AVOL \geq 50\text{K}$ ,  $V_{IO} \leq \pm 2.5\text{mV}$  and  $IBIAS \leq 1.0 \mu\text{A}$  at room temperature.

### Radiation Effects

- (1) Total Dose  
 Little or no effect will be observed at  $1 \times 10^4$  Rad (Si).  $IBIAS$ ,  $I_{IO}$  and  $AVOL$  starts to degrade between  $1 \times 10^4$  and  $1 \times 10^5$  Rad (Si).
- (2) Dose Rate  
 Devices are constructed in DI and consequently are latchup free.



## Test Product Flow

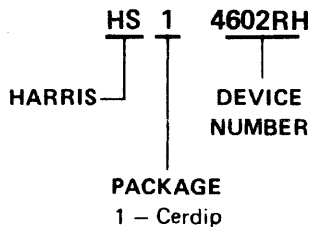
### HARRIS SEMICONDUCTOR PRODUCT FLOW MIL-STD-883, METHOD 5004 CLASS B

#### 100% SCREENING PROCEDURE

1 2 3 4 5 6 7 8 9	SCREEN	MIL-STD-883 METHOD/COND.
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
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7	Burn-In Test	1015, 160 hrs. @ 125°C (or equivalent) (Burn-In circuits enclosed)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

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  - Branding:** All devices are branded with the part number and EIA date code.
  - Aged Products:** Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
  - Additional Requirements:** Sample Group A electrical tests are performed on a lot acceptance basis.

### Ordering Information



### Sales Offices

#### UNITED STATES

##### EASTERN REGION

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
TWX: 710-822-9301

Five Old Concord Road  
Burlington, MA 01803  
(617) 273-1020  
TWX: 710-332-1074

106 Seventh Street  
Garden City, NY 11530  
(516) 747-6776

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1503 South Coast Drive  
Suite 320  
Costa Mesa, CA 92626  
(714) 957-6557  
TWX: 910-595-1533

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Mt. View, CA 94043  
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Minneapolis, MN 55420  
(612) 854-3224  
TWX: 910-576-3418

##### HOME OFFICE

P.O. Box 883  
Melbourne, FL 32901  
(305) 724-7045  
TWX: 510-959-6259

#### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Division  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



# HARRIS

CUSTOM INTEGRATED CIRCUITS DIVISION



*Preliminary*

# HS-508ARH

Radiation Resistant  
8 Channel CMOS Analog Multiplexer  
With Overvoltage Protection

## Features

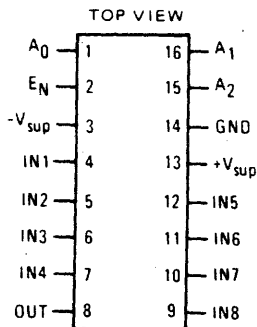
- ANALOG/DIGITAL OVERVOLTAGE PROTECTION
- FAIL SAFE WITH POWER LOSS (NO LATCHUP)
- BREAK-BEFORE-MAKE SWITCHING
- DTL/TTL AND CMOS COMPATIBLE
- ANALOG SIGNAL RANGE  $\pm 15V$
- ACCESS TIME (TYP.) 500ns
- SUPPLY CURRENT AT 1MHz ADDRESS TOGGLE (TYP.) 4mA
- STANDBY POWER (TYP.) 7.5mW
- RADIATION ENVIRONMENT
  - NEUTRON FLUENCE ( $\phi$ )  $1 \times 10^9$  n/cm<sup>2</sup> ( $E \geq 10$ KeV)
  - GAMMA RATE ( $\dot{\gamma}$ )  $1 \times 10^8$  RADs(Si)/s
  - GAMMA DOSE ( $\dot{\gamma}$ )  $1 \times 10^5$  RADs(Si)

## Description

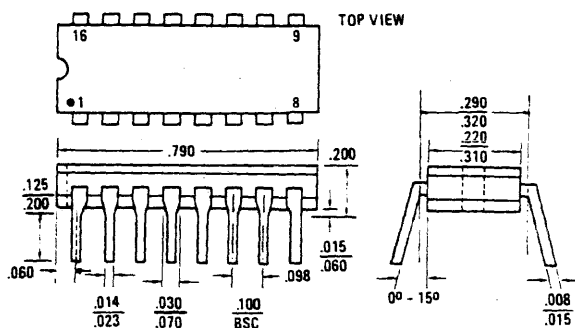
The HS-508ARH is a dielectrically isolated, radiation resistant, CMOS analog multiplexer incorporating an important feature; it withstands analog input voltages much greater than the supplies. This is essential in any system where the analog inputs originate outside the equipment. They can withstand a continuous input up to 10 volts greater than either supply, which eliminates the possibility of damage when supplies are off, but input signals are present. Equally important, it can withstand brief input transient spikes of several hundred volts; which otherwise would require complex external protection networks. Necessarily, ON resistance is somewhat higher than similar unprotected devices, but very low leakage current combine to produce low errors. Reference Application Notes 520 and 521, available from the Analog Products Division of Harris, for further information on the 508A multiplexer in general.

The HS-508ARH has been specifically designed to meet exposure to radiation environments. Operation from  $-55^{\circ}C$  to  $+125^{\circ}C$  is guaranteed.

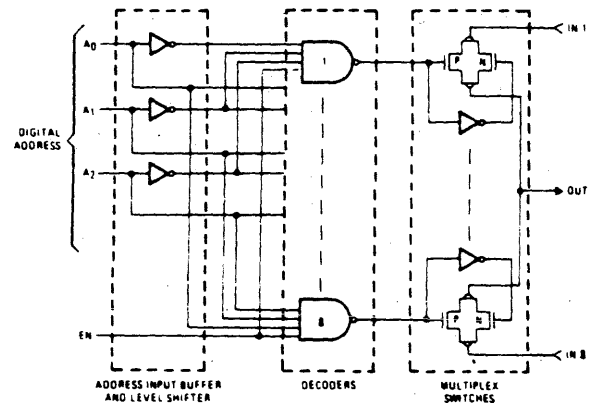
## Pinout



## Package



## Functional Diagram



CAUTION: These devices are sensitive to electrostatic discharge.

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Harris Semiconductor

CUSTOM/SEMICUSTOM



# Specifications HS-508ARH

## ABSOLUTE MAXIMUM RATINGS

Voltage between Supply Pins	40V	Total Power Dissipation*	725 mW
V+ to Ground	20V	Operating Temperature	-55°C to +125°C
VEN, VA, Digital Input Overvoltage:			
VA { VSupply (+) +4V			
VSupply (-) -4V			
Analog Input Overvoltage:		Storage Temperature	
VS { VSupply (+) +20V		-65°C to +150°C	
VSupply (-) -20V			

\*Derate 8mW/°C above tA = 75°C

## ELECTRICAL CHARACTERISTICS (Unless Otherwise Specified)

Supplies = +15V, -15V; VAH (Logic Level High) = +4.0V; VAL (Logic Level Low) = +0.8V  
For Test Conditions, consult Performance Characteristics section.

PARAMETER	TEMP	-55°C to +125°C			UNITS
		MIN	TYP	MAX	
<b>ANALOG CHANNEL CHARACTERISTICS</b>					
*VS, Analog Signal Range	Full	-15		+15	V
*RON, On Resistance (Note 1)	+25°C		1.2	1.5	K Ω
	Full		1.5	1.8	K Ω
*IS(OFF), Off Input Leakage Current	+25°C		0.03		nA
	Full			±50	nA
*ID(OFF), Off Output Leakage Current	+25°C		1.0		nA
	Full			±250	nA
*ID(OFF) with Input Overvoltage Applied (Note 2)	+25°C		4.0		nA
	Full			2.0	μA
*ID(ON), On Channel Leakage Current	+25°C		0.1		nA
	Full			±250	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>					
VAL, Input Low Threshold	(Note 6)	Full		0.8	V
VAH, Input High Threshold		Full	4.0		
*IA, Input Leakage Current (High or Low)	Full			1.0	μA
<b>SWITCHING CHARACTERISTICS</b>					
tA, Access Time	+25°C		0.5	1.0	μs
tOPEN, Break - Before Make Delay	+25°C		80		ns
tON(EN), Enable Delay (ON)	+25°C		300		ns
tOFF(EN), Enable Delay (OFF)	+25°C		300		ns
Settling Time (0.1%)	+25°C		1.2		μs
(0.025%)	+25°C		3.5		μs
"OFF Isolation" (Note 3)	+25°C		65		dB
CS(OFF), Channel Input Capacitance	+25°C		5		pF
CD(OFF), Channel Output Capacitance	+25°C		25		pF
CA, Digital Input Capacitance	+25°C		5		pF
CDS(OFF), Input to Output Capacitance	+25°C		0.1		pF
<b>POWER REQUIREMENTS</b>					
PD, Power Dissipation	Full		7.5		mW
*I+, Current (Note 4)	Full		0.5	2.0	mA
*I-, Current (Note 4)	Full		0.02	1.0	mA
*I+, Standby (Note 5)	Full		0.5	2.0	mA
*I-, Standby (Note 5)	Full		0.02	1.0	mA

## Truth Table

A2	A1	A0	EN	"ON" CHANNEL
X	X	X	L	NONE
L	L	L	H	1
L	L	H	H	2
L	H	L	H	3
L	H	H	H	4
H	L	L	H	5
H	L	H	H	6
H	H	L	H	7
H	H	H	H	8

- NOTES: 1. VOUT = ±10V, IOUT = -100 μA  
 2. Analog Overvoltage = ±33V  
 3. VEN = 0.8V, RL = 1K, CL = 7pF, VS = 3V RMS, f = 500KHz  
 4. VEN = +4.0V  
 5. VEN = 0.8V  
 6. To drive from DTL/TTL Circuits, 1K Ω pull-up resistors to +5.0V supply are recommended

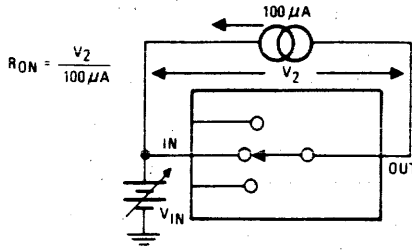
100% Tested for Dash 8 at +25°C and +125°C Only.

# Performance Characteristics and Test Circuits



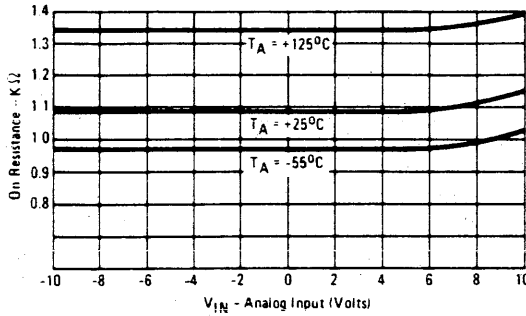
UNLESS OTHERWISE SPECIFIED:  $T_A = 25^\circ\text{C}$ ,  $V_{\text{SUPPLY}} = +15\text{V}$ ,  $V_{\text{AH}} = +4\text{V}$ ,  $V_{\text{AL}} = 0.8\text{V}$

**TEST CIRCUIT NO. 1**

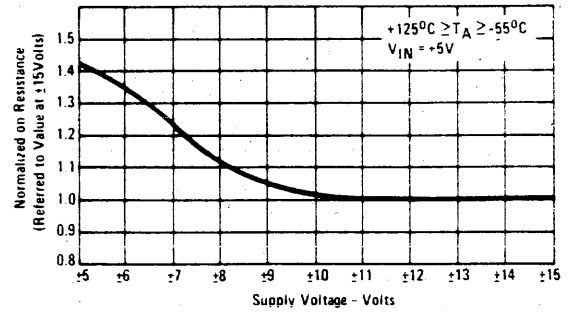


**ON RESISTANCE vs. INPUT SIGNAL LEVEL, SUPPLY VOLTAGE**

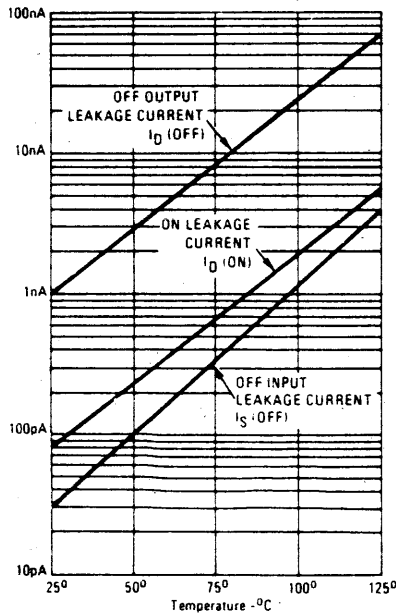
**ON RESISTANCE vs. ANALOG INPUT VOLTAGE**



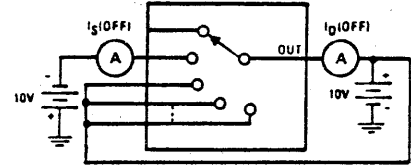
**NORMALIZED ON RESISTANCE vs. SUPPLY VOLTAGE**



## LEAKAGE CURRENT vs. TEMPERATURE

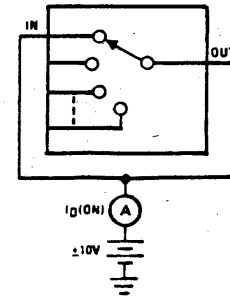


**TEST CIRCUIT OFF LEAKAGE CURRENT vs. TEMPERATURE**

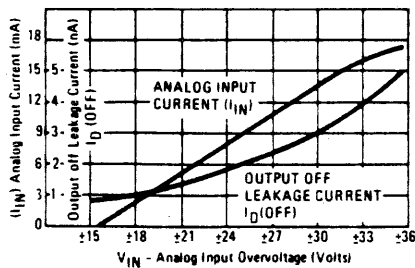


**ON LEAKAGE CURRENT vs. TEMPERATURE**

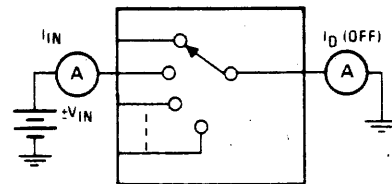
**TEST CIRCUIT NO. 3**

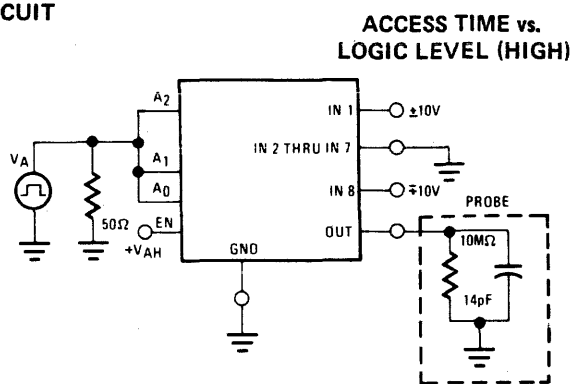
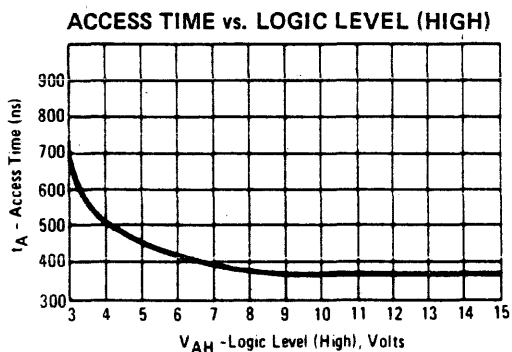
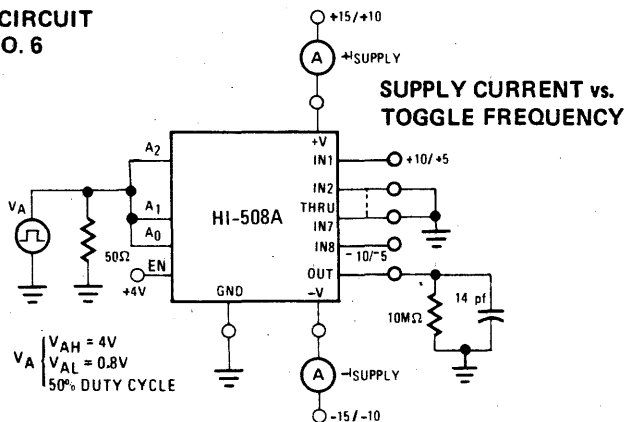
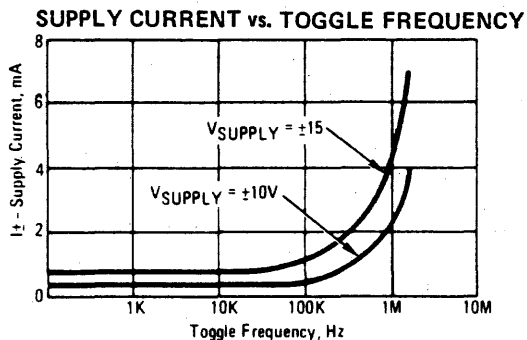
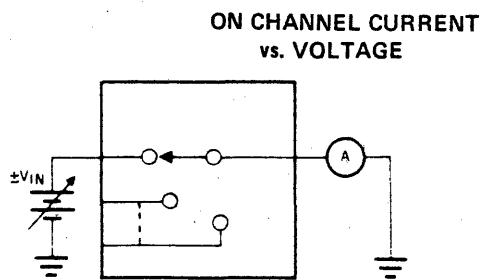
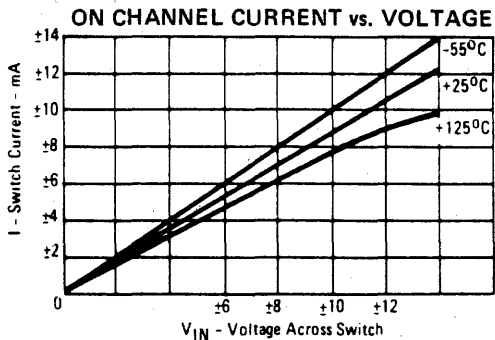


## ANALOG INPUT OVERVOLTAGE CHARACTERISTICS

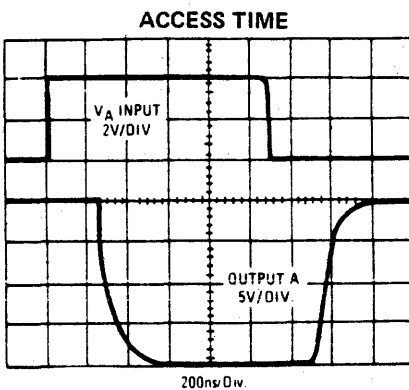
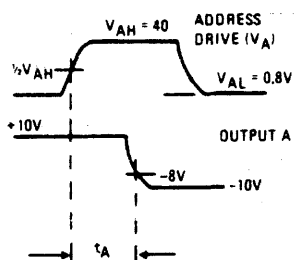


**TEST CIRCUIT NO. 4**

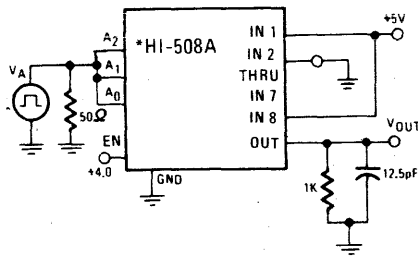
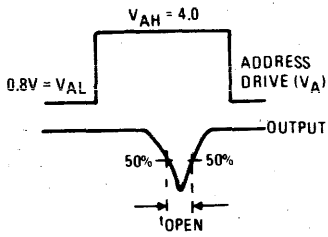




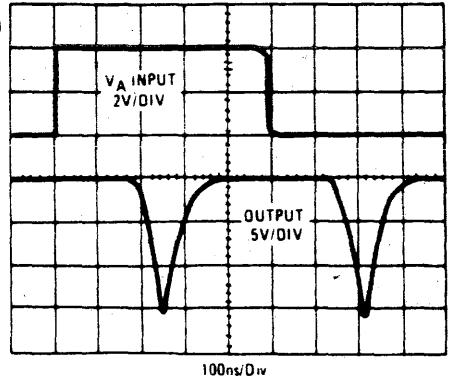
**Switching Waveforms**



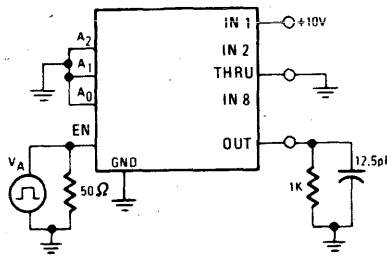
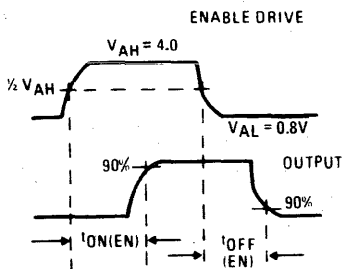
TEST CIRCUIT NO. 8  
BREAK BEFORE MAKE DELAY ( $t_{OPEN}$ )



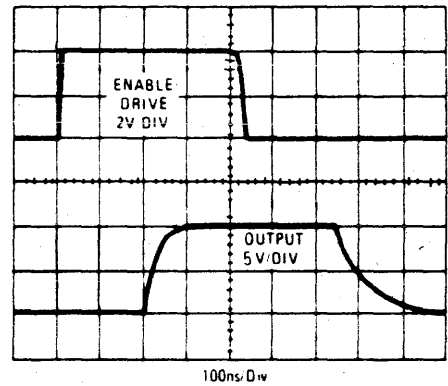
BREAK BEFORE MAKE DELAY ( $t_{OPEN}$ )



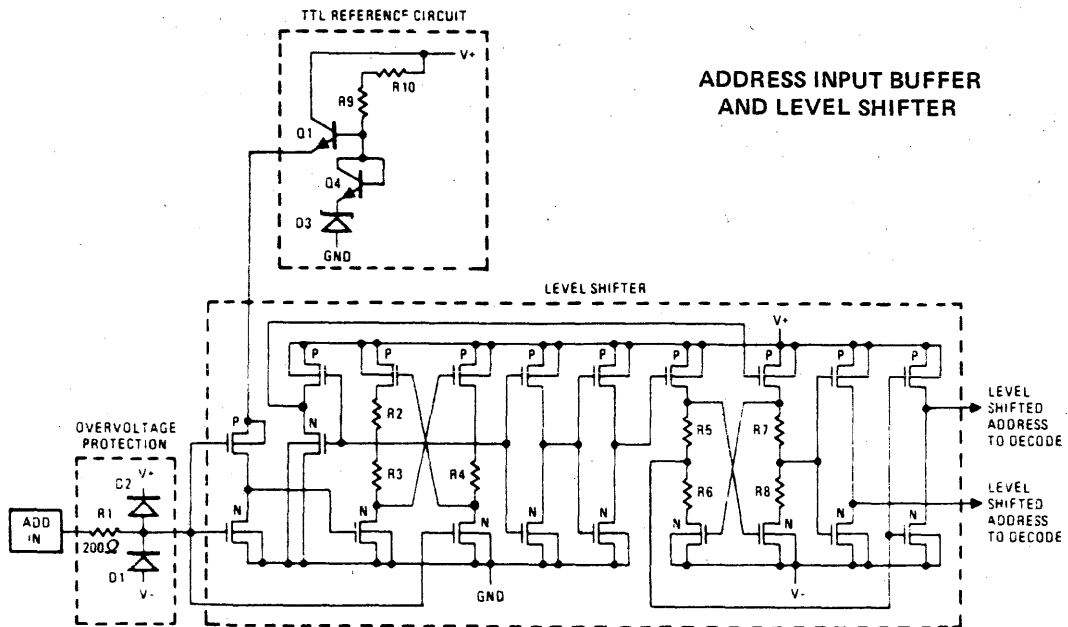
TEST CIRCUIT NO. 9  
ENABLE DELAY ( $t_{ON(EN)}$ ,  $t_{OFF(EN)}$ )



ENABLE DELAY ( $t_{ON(EN)}$ ,  $t_{OFF(EN)}$ )

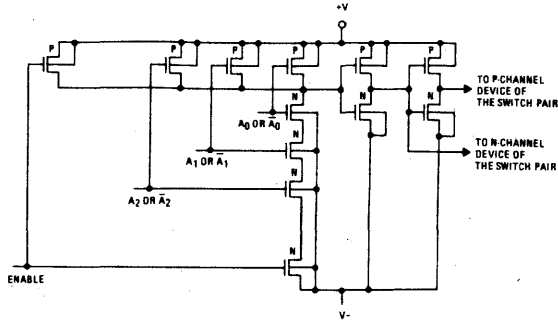


Schematic Diagrams

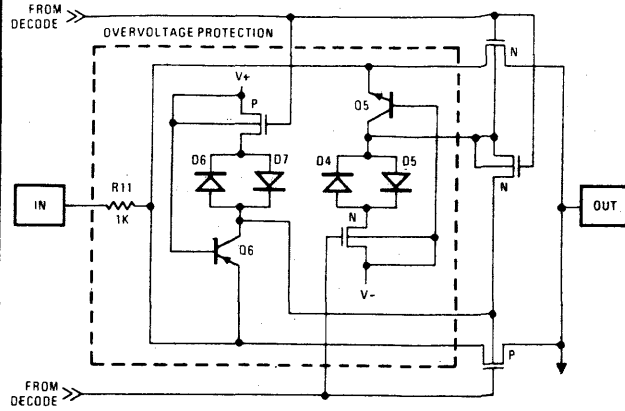


## Schematic Diagrams (continued)

### ADDRESS DECODER



### MULTIPLEX SWITCH



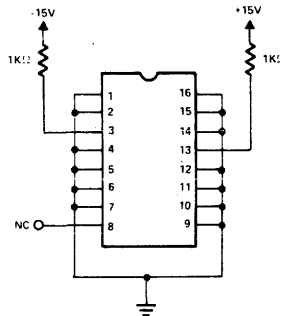
### Radiation Screening Procedure

- (1) Two (2) probed good samples per wafer will be selected from  $\geq 20\%$  of the wafers in a "run". (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $1 \times 10^5$  Rad (Si)  $\pm 10\%$  from a Gamma Cell 220 Cobalt 60 source or equivalent. The devices will be powered in the configuration illustrated with  $V_{SUPPLY} = \pm 15V$ . The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4) On Current Leakage  $I_{D(ON)}$  with  $V_{SUPPLY} = \pm 15V$  will be measured and recorded for each device within one hour after irradiation. The lot will be accepted only if the sample meets the limit of  $I_{D(ON)} \leq 1\mu A$  when tested at  $25^\circ C$ .

### Radiation Effects

- (1) TOTAL DOSE:  
Very little degradation of any of the parameters will be seen up to  $\gamma = 1 \times 10^5$  Rad (Si).
- (2) DOSE RATE:  
The HS508A RH is manufactured in DI, consequently, it is latch up free.

### Irradiation Bias Circuit



### Sales Offices

#### UNITED STATES

##### EASTERN REGION

2600 Virginia Avenue  
Suite 800  
Washington, DC 20037  
(202) 342-3900  
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(305) 724-7045  
TWX: 510-959-6259

#### EUROPEAN

Harris Systems Ltd.  
Semiconductor Programs Divisor  
P.O. Box 27  
145 Farnham Road  
Slough SL1 4XD  
United Kingdom  
Tel: 34666  
TWX: 848174



# HARRIS

CUSTOM INTEGRATED CIRCUITS DIVISION



*Preliminary*

# HS-1840RH

Radiation Resistant  
16 Channel CMOS Analog  
Multiplexer with High-Z Analog  
Input Protection

## Features

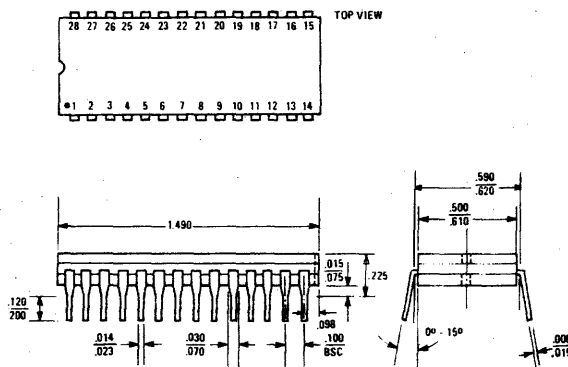
- HIGH ANALOG INPUT IMPEDANCE DURING POWER LOSS (OPEN) 500M $\Omega$
- LOW POWER CONSUMPTION (STANDBY) 600 $\mu$ W
- ACCESS TIME 500ns
- EXCELLENT IN HI-REL REDUNDANT SYSTEMS
- BREAK-BEFORE-MAKE SWITCHING
- NO LATCH-UP
- RADIATION ENVIRONMENT
  - NEUTRON FLUENCE ( $\phi$ ) . . . . .  $1 \times 10^9$  n/cm<sup>2</sup>(E  $\geq$  10KeV)
  - GAMMA RATE ( $\dot{\gamma}$ ) . . . . .  $1 \times 10^8$  RADs (Si)/s
  - GAMMA DOSE ( $\gamma$ ) . . . . .  $2 \times 10^5$  RADs(Si)

## Description

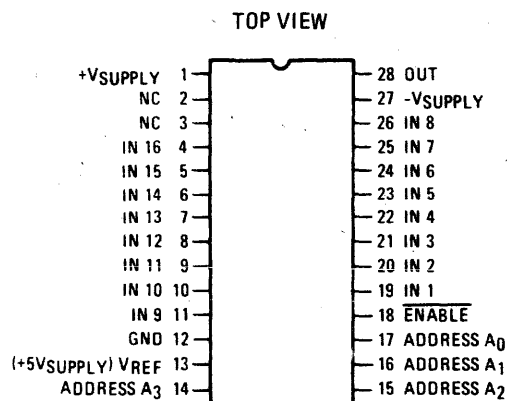
The HS-1840RH is a radiation resistant, monolithic 16 channel multiplexer constructed with the Harris Linear Dielectric Isolation CMOS process. It is designed to provide a high input impedance to the analog source if device power fails (open) or the analog signal voltage inadvertently exceeds the supply rails during powered operation. Excellent for use in redundant applications, since the secondary device can be operated in a standby unpowered mode affording no additional power drain. But more significantly, a very high impedance exists between the active and inactive devices preventing any interaction. One of sixteen channel selection is controlled by a 4-bit binary address plus an Enable-Inhibit input which conveniently controls the ON/OFF operation of several multiplexers in a system. All digital inputs have electrostatic discharge protection.

The HS-1840RH has been specifically designed to meet exposure to radiation environments. It is available in a 28 pin dual-in-line package and is guaranteed operational from  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ .

## Package

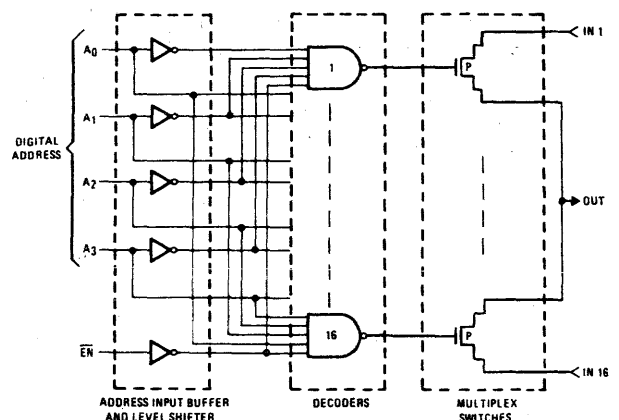


## Pinout



CAUTION: These devices are sensitive to electronic discharge.

## Functional Diagram







# Specifications HS-1840RH

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage Between Pins 1 and 27	+40V	Total Power Dissipation*	1200mW
V <sub>REF</sub> to Ground	+20V	Operating Temperature	-55°C to +125°C
V <sub>EN</sub> , V <sub>A</sub> , Digital Input Overvoltage:		Storage Temperature	-65°C to +150°C
V <sub>A</sub> V <sub>REF</sub> +4V			
Ground -4V			
Analog Input Overvoltage:			
V <sub>S</sub> V <sub>Supply</sub> (+) +20V			
V <sub>Supply</sub> (-) -20V			

\*Derate 8mW/°C above T<sub>A</sub> = +25°C

## ELECTRICAL CHARACTERISTICS

Unless Otherwise Specified:

Supplies = +15V, -15V: V<sub>REF</sub>(Pin 13) = +5V; V<sub>AH</sub>(Logic Level High) = 4.0V; V<sub>AL</sub>(Logic Level Low) = +0.8V  
For Test Conditions, consult Performance Characteristics section.

PARAMETER	TEMP.	-55°C to +125°C			UNITS
		MIN.	TYP.	MAX.	
<b>ANALOG CHANNEL CHARACTERISTICS</b>					
*V <sub>S</sub> , Analog Signal Range	Full	-5		+15	V
*R <sub>ON</sub> , On Resistance (Note 1) V <sub>IN</sub> = +15V	Full		0.5	1.0	KΩ
V <sub>IN</sub> = -5V	Full		2.5	5.0	KΩ
*I <sub>S</sub> (OFF), Off Input Leakage Current	+25°C		0.03		nA
	Full			±100	nA
*I <sub>S</sub> (OFF), with Power Off (Note 8)	Full			±100	nA
*I <sub>O</sub> (OFF), Off Output Leakage Current	+25°C		1.0		nA
	Full			±1000	nA
*I <sub>O</sub> (OFF), or I <sub>S</sub> (OFF) with Input Overvoltage Applied (Note 2)	+25°C		50		nA
	Full			±1000	nA
*I <sub>O</sub> (ON), On Channel Leakage Current	+25°C		1.0		nA
	Full			±1000	nA
<b>DIGITAL INPUT CHARACTERISTICS</b>					
V <sub>AL</sub> , Input Low Threshold   TTL Drive	Full			0.8	V
V <sub>AH</sub> , Input High Threshold   (Note 7)	Full	4.0			V
V <sub>AL</sub>   MOS Drive (Note 3)	+25°C			0.8	V
V <sub>AH</sub>	+25°C	6.0			V
*I <sub>A</sub> , Input Leakage Current (High or Low)	Full			1.0	μA
<b>SWITCHING CHARACTERISTICS</b>					
t <sub>A</sub> , Access Time	+25°C		500	1000	ns
t <sub>OPEN</sub> , Break-Before-Make Delay	+25°C	20	80		ns
t <sub>ON</sub> (EN), Enable Delay (ON)	+25°C		300	1000	ns
t <sub>OFF</sub> (EN), Enable Delay (OFF)	+25°C		300	1000	ns
Settling Time (0.1%)	+25°C		1.2		μs
(0.025%)	+25°C		4.1		μs
"Off Isolation" (Note 4)	+25°C		65		dB
C <sub>S</sub> (OFF), Channel Input Capacitance	+25°C		5		pF
C <sub>O</sub> (OFF), Channel Output Capacitance	+25°C		50		pF
C <sub>A</sub> , Digital Input Capacitance	+25°C		5		pF
C <sub>DS</sub> (OFF), Input or Output Capacitance	+25°C		0.15		pF
<b>POWER REQUIREMENTS</b>					
P <sub>D</sub> , Power Dissipation (Note 5)	+25°C		0.6	15.0	mW
(Note 6)	+25°C		0.6	15.0	mW
*I <sub>+</sub> , Current Pin 1 (Note 5)	Full		0.02	0.5	mA
*I <sub>-</sub> , Current Pin 27 (Note 5)	Full		0.02	0.5	mA
*I <sub>+</sub> , Standby (Note 6)	Full		0.02	0.5	mA
*I <sub>-</sub> , Standby (Note 6)	Full		0.02	0.5	mA

## Truth Table

A <sub>3</sub>	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	EN	"ON" CHANNEL
X	X	X	X	H	NONE
L	L	L	L	L	1
L	L	L	H	L	2
L	L	H	L	L	3
L	L	H	H	L	4
L	H	L	L	L	5
L	H	L	H	L	6
L	H	H	L	L	7
L	H	H	H	L	8
H	L	L	L	L	9
H	L	L	H	L	10
H	L	H	L	L	11
H	L	H	H	L	12
H	H	L	L	L	13
H	H	L	H	L	14
H	H	H	L	L	15
H	H	H	H	L	16

### NOTES:

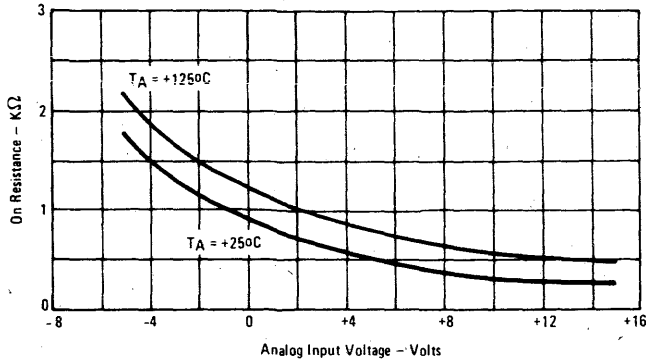
- I<sub>OUT</sub> = 1mA
- Analog Overvoltage = ±20V
- V<sub>REF</sub> = +10V
- V<sub>EN</sub> = 4.0V, R<sub>L</sub> = 1K, C<sub>L</sub> = 7pF, V<sub>S</sub> = 3VRMS, f = 500kHz
- V<sub>EN</sub> = 0.8V
- V<sub>EN</sub> = 4.0V
- To drive from DTL/TTL circuits 1K pull-up resistors to +5.0V supply are recommended
- All supplies (V<sub>+</sub>, V<sub>-</sub>, +5V) and digital inputs (A<sub>0</sub>, A<sub>1</sub>, A<sub>2</sub>, A<sub>3</sub>, EN) opened. Analog input ±10V.

# Performance Characteristics and Test Circuits

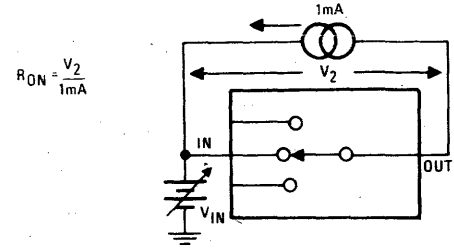


Unless Otherwise Specified:  $T_A = 25^\circ\text{C}$ ,  $V_{\text{SUPPLY}} = \pm 15\text{V}$ ,  $V_{\text{AH}} = +4\text{V}$ ,  $V_{\text{AL}} = 0.8\text{V}$  and  $V_{\text{REF}} = 5\text{V}$ .

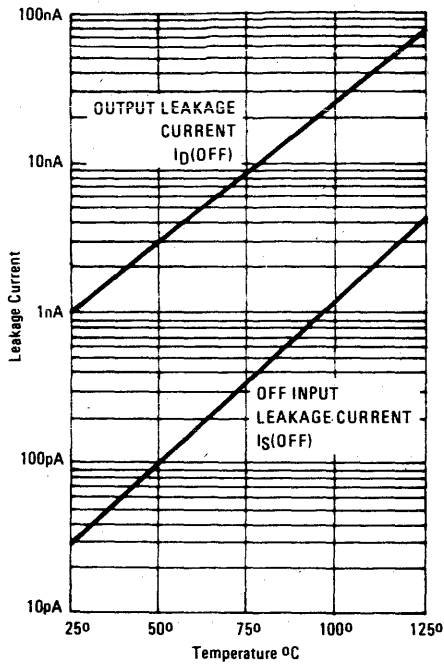
## ON RESISTANCE VS. ANALOG INPUT VOLTAGE



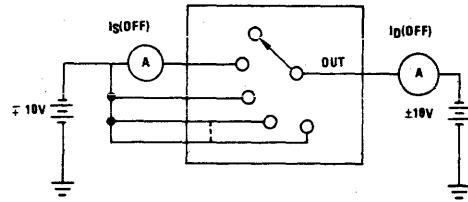
## ON RESISTANCE vs INPUT SIGNAL LEVEL



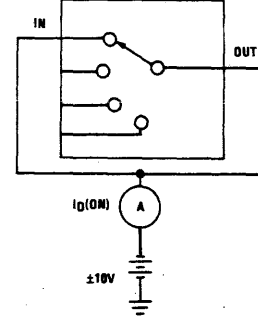
## LEAKAGE CURRENT VS. TEMPERATURE



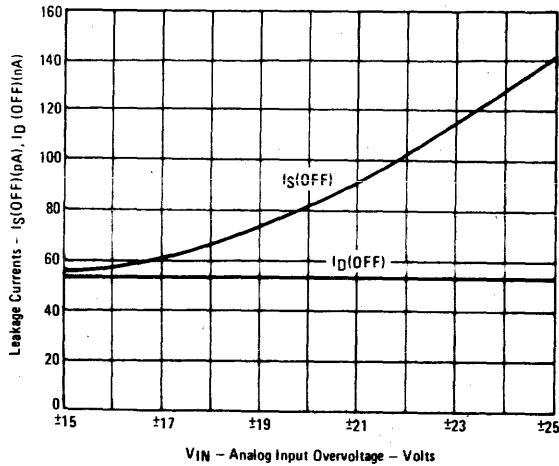
## OFF LEAKAGE CURRENT vs TEMPERATURE



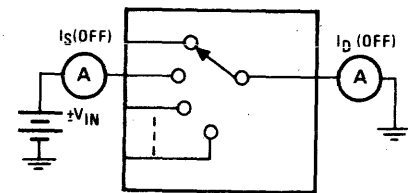
## ON LEAKAGE CURRENT vs TEMPERATURE



## ANALOG INPUT OVERVOLTAGE CHARACTERISTICS

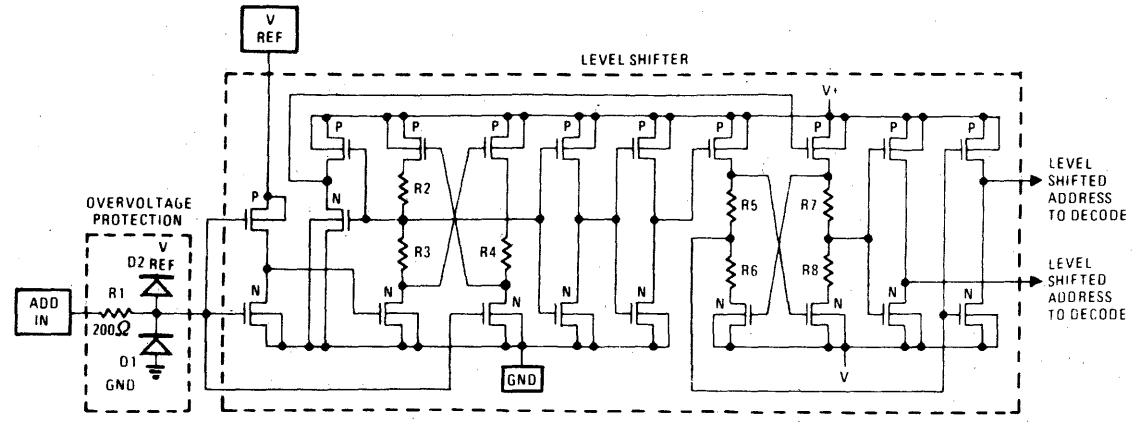


## ANALOG INPUT OVERVOLTAGE CHARACTERISTICS

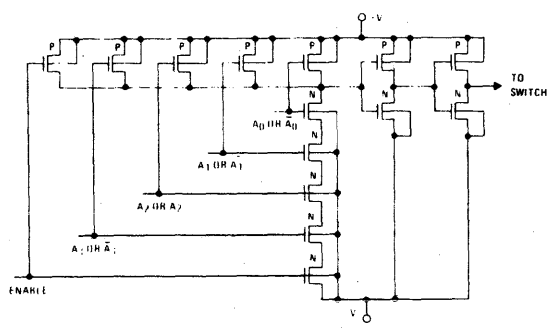


### Schematic Diagrams

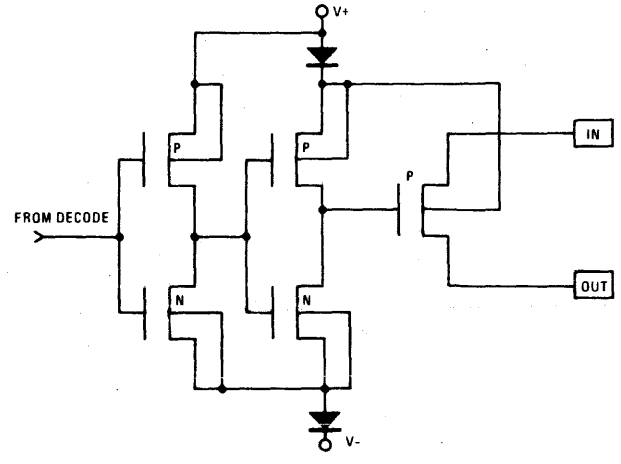
#### ADDRESS AND ENABLE INPUT BUFFER AND LEVEL SHIFTER



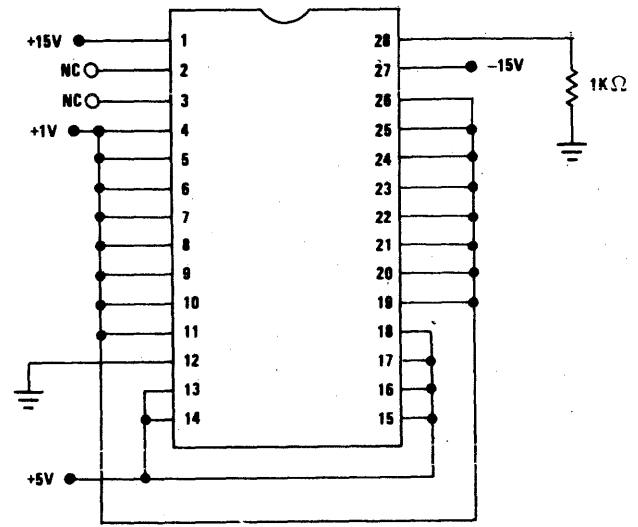
#### ADDRESS DECODER



#### MULTIPLEXER SWITCH



#### HS1840RH IRRADIATION CONFIGURATION





### Radiation Screening Procedure

- (1) Two (2) probed good die per wafer will be selected from  $\geq 20\%$  of the wafers in a run. (All wafers in a "run" will have been processed together through all high temperature processing steps and through metallization.)
- (2) The sample die shall be assembled and tested for functionality.
- (3) The sample devices shall be subjected to a Total Dose Radiation level of  $2 \times 10^5$  Rad (Si)  $\pm 10\%$  from a Gamma Cell 220 Cobalt 60 source or equivalent. The devices will be powered in the configuration illustrated with  $V_{SUPPLY} = \pm 15V$ . The dose rate shall be between 50 rads/sec and 200 rads/sec.
- (4) On Current Leakage  $I_{D(ON)}$  with  $V_{SUPPLY} = \pm 15V$  will be measured and recorded for each device within one hour after irradiation. The lot will be accepted only if the sample meets the limit of  $I_{D(ON)} \leq 1 \mu A$  when tested at 25°C.

### Radiation Effects

- (1) TOTAL DOSE:  
Very little degradation of any of the parameters will be seen up to  $\gamma = 2 \times 10^5$  Rad (Si).
- (2) DOSE RATE:  
The HS-1840RH is manufactured in DI, consequently, it is latch up free.

### Product Test Flow

**HARRIS SEMICONDUCTOR PRODUCT FLOW**  
**MIL-STD-883, METHOD 5004 CLASS B**

**100% SCREENING PROCEDURE**

	SCREEN	MIL-STD-883 METHOD/COND:
1	Internal Visual	2010 Cond. B.
2	Stabilization Bake	1008 Cond. C (24 hrs. minimum)
3	Temperature Cycling	1010 Cond. C
4	Constant Acceleration	2001 Cond. E; Y1 plane
5	Seal: A Fine B Gross	1014 Cond. A or B 1014 Cond. C2
6	Initial Electrical	Harris Specifications
7	Burn-In Test	1015, 160 hrs. 125°C (or equivalent)
8	Final Electrical 100% go-no-go	Tested at Worst Case Operating Conditions
9	External Visual	2009 Sample Inspection

Note:

- Traceability: All devices are assigned date code identification that provides traceability back to the inspection lot.
- Branding: All devices are branded with the part number and EIA date code.
- Aged Products: Product that has been held for more than 24 months will be reinspected prior to shipment to group A inspection requirements.
- Additional Requirements: Sample Group A electrical tests are performed on a lot acceptance basis.

Harris Semiconductor

CUSTOM/SEMICUSTOM



HGA-C00600  
HGA-C01200  
HGA-C02500

CMOS Gate Array

Features

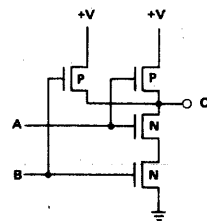
- 3 MICRON CMOS TECHNOLOGY
- 600, 1200, AND 2500 GATE ARRAYS
- 3ns TYPICAL GATE DELAY
- 50 MHz TOGGLE FREQUENCY
- 74LS- SSI, MSI LIBRARY IMPLEMENTED
- ADVANCED CAD TELEDESIGN™ SOFTWARE SUPPORT
- COMPATIBLE WITH DIAL-A-CHIP™ DESIGN AUTOMATION SYSTEM
- INDIVIDUALLY PROGRAMMABLE I/O BUFFERS
- BI-DIRECTIONAL AND THREE STATE I/O
- TTL AND CMOS COMPATIBILITY
- COMMERCIAL TEMPERATURE RANGE
- MILITARY TEMPERATURE RANGE
- MULTIPLE PACKAGE OPTIONS

CMOS Gate Array Family

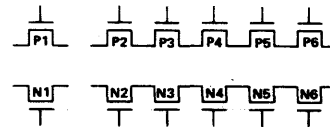
PART NO.	TRAN-SISTOR	GATE EQUIV.	DIE SIZE (MILS)	I/O PINS	ROWS/ CELLS
HGA-C00600	2592	648	198 X 165	54	6 X 36
HGA-C01200	5184	1296	257 X 198	78	12 X 36
HGA-C02500	10,080	2520	272 X 287	100	14 X 60

\* One gate equivalent is equal to one 2-input nand.

TWO-INPUT NAND 1 GATE EQUIVALENT



BASIC CELL



Description

Harris Custom Integrated Circuits Division offers a complete family of custom and semicustom logic. The HGA-C00600, HGA-C01200 and HGA-C02500 are three CMOS gate arrays manufactured using the HARRIS state-of-the-art SAJI IV process. This local oxidation process offers 3 μm channel lengths and 2 μm minimum features. Typical propagation gate delay is 3ns with 50MHz toggle frequencies possible. Personalization is accomplished by patterning two levels of interconnection on three mask layers: polysilicon, contacts and metal.

The entire process is fully supported by advanced CAD Teledesign™\* software. With the HARRIS Dial-A-Chip™\*\* design automation system a customer chooses the amount of his participation in the design and development of his circuit. Using the HARRIS Teledesign™ timeshare system, logic description, simulation, and artwork editing can all be performed by the customer in his own office via a digital data communications link or if he so chooses, he may use the Customer Design Center in Melbourne.

The HARRIS system minimizes the amount of design and logic coding by offering a library of 74LS equivalent SSI and MSI function designs. There is no need to implement large MSI functions with primitive gates. Each function has a complete data sheet specifying propagation delays and A. C. parameters. A final software program verifies that the original logic description matches the finished artwork.

Each input-output buffer can be individually specified to be input, output, three-state, or bi-directional. CMOS or TTL compatibility can also be specified. The large number of I/O buffers keeps the input/output to gate ratio high insuring that the design is not pin limited.

Transition from gate arrays to standard cell circuits is very simple. The same logic description files are used as input to the Dial-A-Chip™ software. Both utilize the same SSI, MSI 74LS library.

\*Trademark of Harris Corporation  
\*\*Servicemark of Harris Corporation



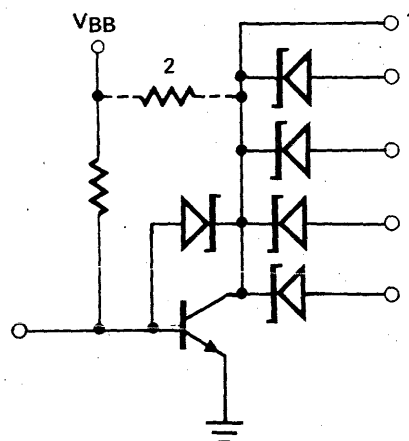
*Advance*

HGA-B01500  
HGA-B02500

ASTL Gate Array

### Features

- 3 MICRON TECHNOLOGY
- 1500, AND 2500 GATE ARRAYS
- 1.0ns TYPICAL GATE DELAY
- 150MHz TOGGLE FREQUENCY
- 74LS-SSI, MSI LIBRARY
- ADVANCED CAD TELEDISIGN™ SOFTWARE SUPPORT
- COMPATIBLE WITH DIAL-A-CHIP™ DESIGN AUTOMATION SYSTEM
- INDIVIDUALLY PROGRAMMABLE I/O BUFFERS
- BI-DIRECTIONAL AND THREE STATE I/O
- TTL COMPATIBILITY
- COMMERCIAL TEMPERATURE RANGE
- MILITARY TEMPERATURE RANGE
- MULTIPLE PACKAGE OPTIONS
- 265 MICROWATTS PER GATE TYPICAL



Notes: 1. Ohmic contact for Common Base connections.  
2. Collector Resistor Option for high Fan-Out.  
Refer to appropriate HBAXXX Data Sheets.

### Description

Harris Custom Integrated Circuits Division offers a complete family of custom and semi-custom logic. The HGA-B01500 and HGA-B02500 are ASTL gate arrays manufactured using the HARRIS state of the art advanced schottky transistor logic process. The process offers 10 $\mu$ m metal pitch and 3 $\mu$ m minimum features. Typical propagation gate delay is 1.0ns with 150MHz toggle frequencies possible. Personalization is accomplished by patterning two levels of aluminum interconnection on four mask layers.

The entire process is fully supported by advanced CAD Teledesign™\* software. With the HARRIS Dial-A-Chip™\*\* design automation system a customer chooses the amount of his participation in the design and development of his circuit. Using the HARRIS Teledesign™ software, logic description, simulation, and artwork editing can all be performed by the customer in his own office via a digital data communications link or if he so chooses, he may use the Customer Design Center in Melbourne.

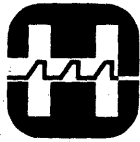
The HARRIS system minimizes the amount of design and logic coding by offering a library of 74LS equivalent SSI and MSI function designs. There is no need to design large MSI functions with primitive gates. Each function has a complete data sheet specifying propagation delays and A.C. parameters. A final software program verifies that the original logic description matches the finished artwork.

Each input-output buffer can be individually specified to be input, output, three-state, or bi-directional. The large number of I/O buffers keeps the gate to input/output ratio low insuring that the design is not pin limited.

Transition from gate arrays to a standard cell design is very simple. The same logic description files are used as input to the Teledesign™ software. Both utilize the same SSI, MSI 74LS library, functions.

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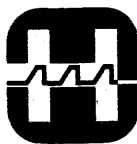


## Semicustom Capabilities

# HCA SSI/MSI LIBRARY

For use with HARRIS CMOS Gate Array  
and Standard Cell Family

HCA000XB	DUAL 2 INPUT NAND
HCA000IB	DUAL 2 INPUT NAND WITH INVERTER
HCA002XB	DUAL 2 INPUT NOR
HCA002IB	DUAL 2 INPUT NOR WITH INVERTER
HCA004XB	TRIPLE INVERTER
HCA007XB	HIGH FANOUT BUFFER (5 pF DRIVE)
HCA008XB	2 INPUT AND
HCA010XB	3 INPUT NAND
HCA011XB	3 INPUT AND
HCA020XB	4 INPUT NAND
HCA021XB	4 INPUT AND
HCA027XB	3 INPUT NOR
HCA029XB	6 INPUT NAND
HCA030XB	8 INPUT NAND
HCA032XB	2 INPUT OR
HCA051XB	4 INPUT AND-OR-INVERT GATE
HCA073XB	JK FLIP FLOP WITH RESET
HCA074RB	D FLIP FLOP WITH RESET
HCA074SB	D FLIP FLOP WITH SET
HCA074XB	D FLIP FLOP WITH SET AND RESET
HCA074RB	D FLIP FLOP WITH RESET—NEGATIVE EDGE TRIGGERED
HCA074SB	D FLIP FLOP WITH SET—NEGATIVE EDGE TRIGGERED
HCA074XB	D FLIP FLOP WITH SET AND RESET—NEGATIVE EDGE TRIGGERED
HCA075XB	QUAD D TYPE LATCH WITH INDIVIDUAL RESETS
HCA083SB	BCD FULL ADDER
HCA085XB	4-BIT MAGNITUDE COMPARATOR
HCA086XB	2 INPUT EXCLUSIVE OR
HCA138XB	3 TO 8 LINE DECODER
HCA139XB	2 TO 4 LINE DECODER
HCA152XB	8 TO 1 DATA SELECTOR
HCA157XB	QUAD 2 TO 1 DATA SELECTOR
HCA161XB	4-BIT SYNCHRONOUS BINARY COUNTER WITH ASYNCHRONOUS RESET
HCA163XB	4-BIT SYNCHRONOUS BINARY COUNTER WITH SYNCHRONOUS RESET
HCA164XB	8-BIT PARALLEL OUTPUT SHIFT REGISTER
HCA165XB	8-BIT PARALLEL LOAD SHIFT REGISTER
HCA173XB	4-BIT D REGISTER WITH THREE STATE OUTPUTS
HCA175XB	QUAD D FLIP FLOP WITH RESET
HCA180XB	9-BIT PARITY GENERATOR
HCA192XB	4-BIT SYNCHRONOUS UP DOWN DECADE COUNTER
HCA193XB	4-BIT SYNCHRONOUS UP DOWN BINARY COUNTER
HCA194XB	4-BIT SYNCHRONOUS LOAD BI-DIRECTIONAL SHIFT REGISTER
HCA225XB	EXPANDABLE 4 WORD BY 1 BIT FIFO
HCA240XB	OCTAL INVERTING THREE STATE BUFFER
HCA244XB	OCTAL NON-INVERTING THREE STATE BUFFER
HCA257XB	QUAD 2 TO 1 LINE DATA SELECTOR
HCA273XB	OCTAL D FLIP FLOP WITH COMMON CLOCK
HCA283XB	4-BIT BINARY FULL ADDER WITH FAST CARRY
HCA352XB	DUAL 4 TO 1 LINE DATA SELECTOR
HCA377XB	OCTAL D FLIP FLOP WITH ENABLE
HCA393XB	4-BIT BINARY RIPPLE COUNTER
HCA645XB	OCTAL BUS TRANSCEIVER
HCA1000B	15:1 RATIO PULL UP RESISTOR WITH ENABLE



## Semicustom Capabilities

### HCA SSI/MSI LIBRARY

For use with HARRIS CMOS Gate Array  
and Standard Cell Family

#### GATE ARRAY I/O BUFFERS

HCA900XB	CMOS INPUT BUFFER
HCA901XB	CMOS INPUT BUFFER WITH 18K PULL UP RESISTOR
HCA910XB	TTL INPUT BUFFER
HCA911XB	TTL INPUT BUFFER WITH 18K PULL UP RESISTOR
HCA950XB	3.2 MA OUTPUT BUFFER
HCA951XB	3.2 MA OUTPUT BUFFER—OPEN DRAIN P-CHANNEL
HCA952XB	3.2 MA OUTPUT BUFFER—OPEN DRAIN N-CHANNEL
HCA960XB	3.2 MA OUTPUT BUFFER—THREE STATE
HCA970XB	3.2 MA BI-DIRECTIONAL BUFFER—CMOS INPUT
HCA971XB	3.2 MA BI-DIRECTIONAL BUFFER—CMOS INPUT WITH 18K PULL UP
HCA980XB	3.2 MA BI-DIRECTIONAL BUFFER—TTL INPUT
HCA981XB	3.2 MA BI-DIRECTIONAL BUFFER—TTL INPUT WITH 18K PULL UP RESISTOR
HCAVSSXB	PLACEABLE VSS I/O PAD
HCAVIA	VIA CELL

#### STANDARD CELL I/O BUFFERS

HCC900XB	CMOS INPUT BUFFER
HCC901XB	CMOS INPUT BUFFER WITH 18K PULL UP
HCC905XB	CMOS INPUT BUFFER—INVERTING
HCC906XB	CMOS INPUT BUFFER—INVERTING WITH PULL UP
HCC908XB	SCHMITT TRIGGER INPUT BUFFER
HCC910XB	TTL INPUT BUFFER
HCC911XB	TTL INPUT BUFFER WITH 18K PULL UP
HCC915XB	TTL INPUT BUFFER WITH 18K PULL UP INVERTING
HCC950XB	3.2 MA OUTPUT BUFFER
HCC951XB	3.2 MA OUTPUT BUFFER WITH OPEN DRAIN P-CHANNEL
HCC952XB	3.2 MA OUTPUT BUFFER WITH OPEN DRAIN N-CHANNEL
HCC955XB	8.0 MA OUTPUT BUFFER
HCC960XB	3.2 MA OUTPUT BUFFER THREE STATE
HCC965XB	8.0 MA OUTPUT BUFFER THREE STATE
HCC970XB	3.2 MA BI-DIRECTIONAL WITH CMOS INPUT
HCC971XB	3.2 MA BI-DIRECTIONAL CMOS INPUT AND 18K PULL UP
HCC975XB	8.0 MA BI-DIRECTIONAL WITH CMOS INPUT
HCC976XB	8.0 MA BI-DIRECTIONAL CMOS INPUT AND 18K PULL UP
HCC980XB	3.2 MA BI-DIRECTIONAL TTL INPUT
HCC981XB	3.2 MA BI-DIRECTIONAL TTL INPUT AND 18K PULL UP
HCC985XB	8.0 MA BI-DIRECTIONAL TTL INPUT
HCC986XB	8.0 MA BI-DIRECTIONAL TTL INPUT AND 18K PULL UP
HCC990XB	UNBUFFERED INPUT WITH STATIC PROTECTION
HCCVDDYB	VDD SUPPLY PAD
HCCVSSYB	VSS SUPPLY PAD





# HARRIS

## HSC-CXXXXX

### Standard Cell

#### Features

- 3 MICRON CMOS TECHNOLOGY
- VARIABLE DIE SIZE
- 3ns TYPICAL GATE DELAY
- 50MHz TOGGLE FREQUENCY
- 74LS-SSI, MSI LIBRARY IMPLEMENTED
- ADVANCED CAD TELEDESIGN™\* SOFTWARE SUPPORT
- VARIABLE DIE SIZE
- INDIVIDUALLY PROGRAMMABLE I/O BUFFERS
- COMPATIBLE WITH DIAL-A-CHIP™\*\* DESIGN AUTOMATION SYSTEM
- BI-DIRECTIONAL AND THREE STATE I/O
- TTL AND CMOS COMPATIBILITY
- COMMERCIAL TEMPERATURE RANGE
- MILITARY TEMPERATURE RANGE
- MULTIPLE PACKAGE OPTIONS

#### Description

Harris Custom Integrated Circuits Division offers a complete family of custom and semicustom products. The HARRIS Standard Cell product is a very cost effective alternative to gate arrays or full custom. The standard cell circuit is manufactured using the HARRIS state-of-the-art SAJI IV local oxidation process. The process offers 3  $\mu$ m channel lengths with typical gate delays of 3ns and toggle frequencies of 50MHz. Each input-output buffer can be individually programmed as input, output, three-state or bi-directional. CMOS and TTL compatibility can be specified. The same SSI MSI 74LS functions used to design the gate array are used for the standard cell circuits.

The logic description and simulator functions are identical for both standard cell circuits and gate arrays. However, the standard cell circuit dimensions can change dynamically for each individual circuit. Where the gate array is a fixed number of gates and fixed chip size, the standard cell uses only enough silicon to fully implement the desired logic function. Typically a standard cell chip is 30% smaller than a fully utilized gate array offering a cost savings benefit to those customers with medium to high volumes.

The standard cell chip does require a complete set of masks rather than the three required by the gate array. The standard cell circuit is fully supported by the HARRIS Teledesign™ software and the HARRIS Dial-A-Chip™ timeshare service. This allows the customer to perform as much of the design and development as he desires from his own office with a data terminal.

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enable you to convert your existing discrete or SSI/MSI circuitry to a cost-effective, compact, custom I.C. The tables below list the Monochips you can choose from.

### LINEAR MONOCHIPS

Designation	Technology	Size-Mils	Components	Voltage	Pads	Integration Fee
MOA	Bipolar	71 x 81	260	20V max	16 max	\$3700
MOB	"	81 x 81	300	20V max	24 max	\$3700
MOC	"	51 x 56	110	20V max	14 max	\$2800
MOD	"	80 x 80	194	36V max	16 max	\$3700
MOE	"	70 x 70	187	20V max	18 max	\$3100
MOF	"	91 x 110	437	20V max	24 max	\$4500
MOG	"	75 x 78	310	20V max	18 max	\$3400
MOH	"	77 x 88	382	20V max	18 max	\$3400
MOJ	"	61 x 65	170	20V max	18 max	\$2800
MOL	"	81 x 100	416	20V max	24 max	\$3900
MOM	"	101 x 151	812	20V max	28 max	\$7000
MON	"	123 x 157	1070	20V max	40 max	\$9000
MOP	"	90 x 121	623	20V max	24 max	\$5000
MOQ	"	73 x 76	304	20V max	18 max	\$3400
MLA	CMOS	136 x 185	1695	3 to 15V	42 max	\$8500

### DIGITAL MONOCHIPS

Designation	Technology	Size-Mils	Logic Cells	Gates	Gate Delay	Gate Power	Voltage	Pads	Integration Fee
MUA	Bipolar	131 x 131	225	450	25ns	2.0mW	*1 to 1.5V	40 max	\$6500
MUB	"	"	"	"	200ns	0.2mW	"	"	"
MUC	"	"	"	"	10ns	2.5mW	"	"	"
MSA	"	157 x 157	"	"	8ns	1mW	5V max	"	\$7000
MSB	"	200 x 186	440	880	"	"	"	52 max	\$10000
MSC	"	191 x 205	990	1980	"	"	"	64 max	\$15000
MCA	CMOS	112 x 124	112	140	13ns	<5μW**	3 to 15V	32 max	\$5500
MCB	"	136 x 136	162	200	"	"	"	38 max	\$5500
MCC	"	136 x 174	216	270	"	"	"	44 max	\$6500
MCD	"	174 x 174	224	440	"	"	"	52 max	\$7000
MCE	"	136 x 174	160	340	"	"	"	44 max	\$6500
MCF	"	195 x 197	504	630	"	"	"	60 max	\$8000
MCG	"	219 x 216	640	800	"	"	"	68 max	\$9000
MPA	Si-CMOS	135 x 156	408	510	3.5ns	.14mW***	7V max	60 max	\$7000
MPB	"	166 x 207	800	1000	"	"	"	78 max	\$11000
MPC	"	196 x 239	1200	1500	"	"	"	95 max	\$13000
MPD	"	234 x 251	1600	2000	"	"	"	109 max	\$15000
MPE	"	90 x 112	120	150	"	"	"	36 max	\$5500

\*1 volt requires close work with Interdesign. \*\*Frequency dependent, gate power at low frequency.

\*\*\*Bipolar power at max frequency, CMOS at max frequency.

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# The IGC10000 Series CMOS Gate Arrays

## FEATURES

- Complexity from 408 to 1500 Equivalent 2-input Gates
- Mature Silicon Gate CMOS Technology
  - Low development cost
  - 3.3 to 9V nominal power supply range  $\pm 10\%$
  - Full CMOS temperature range:  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$
  - Resistance to latch-up and electrostatic discharge
- Extensive Macro Cell Library
  - Numerous combinational and sequential macros
  - Facilitates 7400 and 4000-based designs
  - TTL or CMOS compatible I/O
  - Analog capability
- Fully Integrated CAD Software Support
  - Highly efficient auto-routing capability
  - Layout fully verified against input logic
  - Accurate post-layout simulation with calculated RC delays
  - Automatic test code conversion

## GENERAL DESCRIPTION

An IGC10000 Gate Array is a matrix of identical cells, each containing 3 uncommitted N-P transistor pairs. Large numbers of identical arrays are prefabricated and stockpiled. A particular circuit is constructed from a prefabricated array by specifying the interconnections among the transistors within and between cells on the final metal layer. Because all except the final metal layer are prefabricated, the cost advantages of mass production can be realized even for low-volume applications. In addition, prefabrication provides a saving in both design and manufacturing time; in some cases customers can receive prototype chips in as few as 6 weeks after initiation of the project.

In most cases IGC10000 gate arrays are processed with one mask step (a customized metal mask along with a standardized contact mask). For some analog applications or where more routing flexibility is needed, users have the option of programming the contact mask in addition to the metal mask.

## THE IGC10000 FAMILY OF GATE ARRAYS

Figure 1 shows a structural representation of an IGC10000 Gate Array. Each rectangle in the body of the matrix represents an array cell; the rectangles

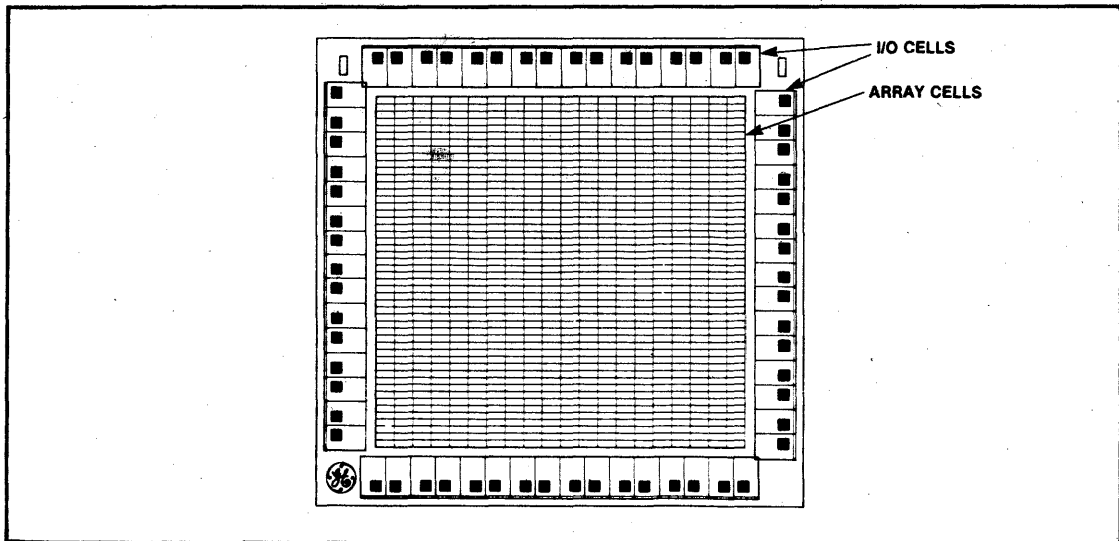


Figure 1. Gate Array Configuration



along each of the four sides of the chip represent I/O cells. Table 1 lists the members of the IGC10000 Gate Array family with their capacities and cell counts.

Table 1. The IGC10000 Gate Array Family

Part No.	Equivalent 2-Input Gates	Number of Array Cells	Bonding Pads and I/O Cells
IGC10408	408	272	34
IGC10756	756	504	44
IGC11500	1500	1000	62

## TECHNOLOGY

IGC10000 gate arrays are fabricated using Intersil's high performance selectively oxidized 4-micron silicon gate CMOS process with single-layer metal interconnect. Wafers are processed using state-of-the-art processing technology with these features:

- Positive photoresist
- 1-1 scanning projection lithography
- Polysilicon, nitride and silicon dioxide plasma etching
- Ion implantation for source/drain doping and threshold adjustment
- Sputter metal deposition
- Industry standard oxidation and diffusion techniques
- Nitride and polysilicon Low Pressure Chemical Vapor Deposition (LPCVD)

## ARRAY CELLS

An array cell, shown in topographical form in Figure 2, consists of three complementary transistor pairs and five strips of polysilicon (called crossunder strips) for making horizontal interconnections among array cells. Power ( $V_{DD}$ ) and ground ( $V_{SS}$ ) buses run vertically as shown. Metal strips (not shown) on top

of the crossunder strips make the vertical interconnections within the array. The top and bottom polysilicon strips (called feed-throughs) are used for feeding horizontal connections through the array cell beneath the power and ground buses.

Figure 3 shows the circuit diagram for the array cell. Small hollow circles represent possible connection points;  $V_{SS}$  and  $V_{DD}$  metal strips are shown as dotted lines.

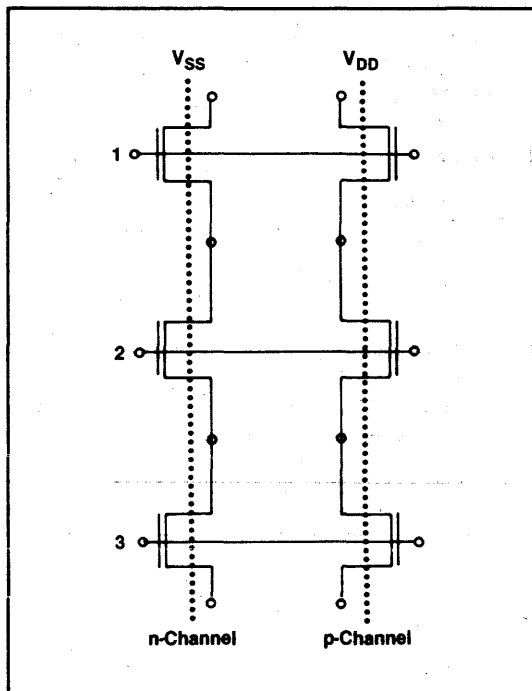


Figure 3. Schematic Diagram of the Array Cell

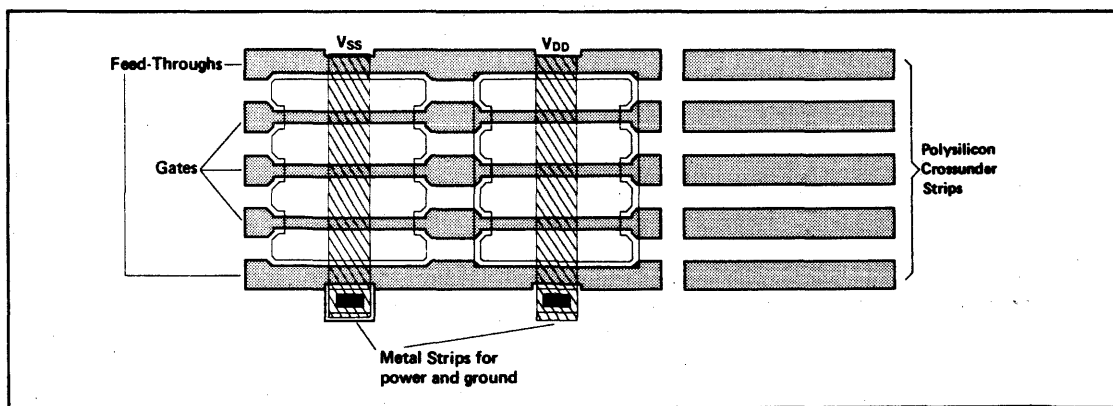


Figure 2. Topography of the Array Cell

# IGC10000



## I/O CELLS

I/O cells are used to interface the array with external circuitry. Each I/O cell consists of an array of transistors of varying sizes, and allows construction of normal digital I/O interface circuits as well as analog circuitry of simple to moderate complexity.

I/O cells have these features:

- Protection against electrostatic discharge (ESD)
- Logic level translation (CMOS-to-TTL and TTL-to-CMOS)
- Bonding pad for connecting the cell to its corresponding package pin
- Ratioed transistors for analog implementation

The output drive capability of a single output buffer is one TTL load. Applications that require additional drive capability can be handled by using multiple I/O cells in parallel. (In a typical application, not all available I/O cells are needed for external connections; unused cells will thus usually be available to provide added drive capability where needed.)

## MACROS

A macro is a physical implementation of a functional block and is realized by interconnections among transistors in one or more array cells. For example, the NOR function is constructed by connecting two p-channel transistors in series to  $V_{DD}$  and two n-channel transistors in parallel to  $V_{SS}$ , as shown in the topographical and schematic diagrams, Figures 4 and 5.

Designers implement their circuits by selecting and interconnecting the macros in the Macro Library, listed in Table 2.

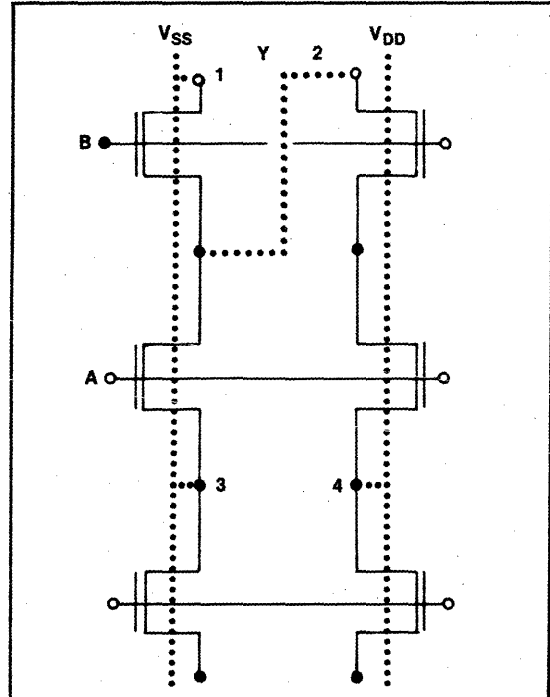


Figure 5. Schematic Diagram for the 2-Input NOR

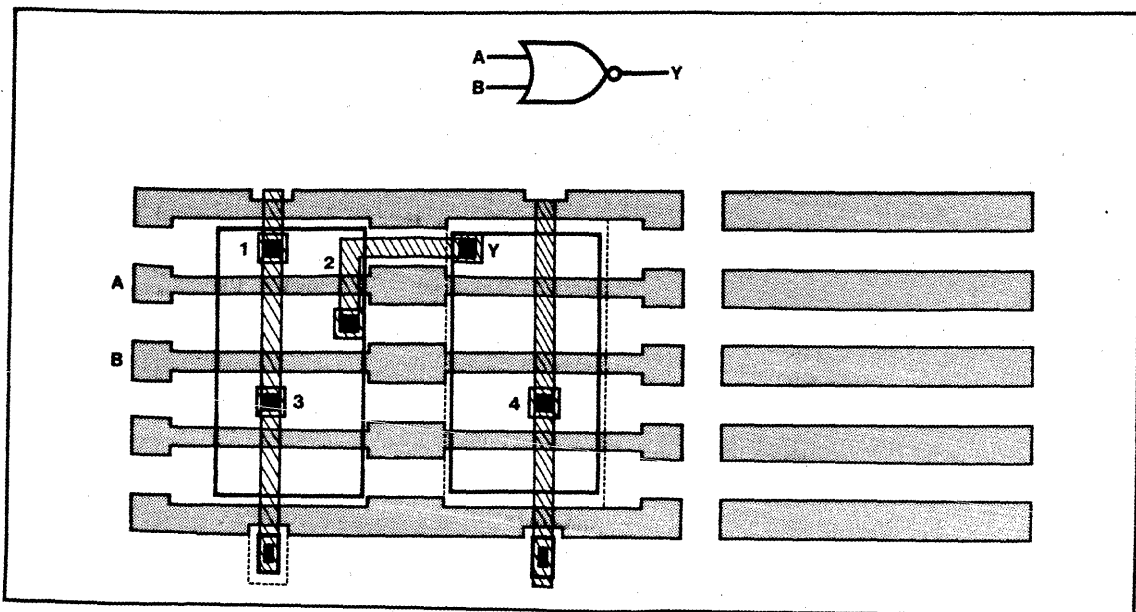


Figure 4. Interconnect Pattern for the 2-Input NOR

Table 2. IGC10000 MACRO Library

IGC10000 Combinational Macros				
Type	Description			
NOR/NAND	2-input NAND 3-input NAND 4-input NAND 2-input NOR 3-input NOR 4-input NOR			
XOR/XNOR	2-input XOR 2-input XNOR			
Adder	One-bit full adder			
Buffers and Inverters	1X inverter 2X inverter 3X inverter 4X inverter			
Multifunction	2-1 AND-OR invert 2-2-2 AND-OR invert 3-2 AND-OR invert 2-1 OR-AND invert			
Multiplexer	2 to 1 multiplexer			
Schmitt Trigger	Schmitt Trigger			
Transmission Gate	Buffered transmission gate Unbuffered transmission gate			
Tristate Control	Tristate control			
IGC10000 Sequential Macros		Set	Reset	Jam Load
D Flip-Flops	D flip-flop with reset	—	Yes	—
	D flip-flop with set	Yes	—	—
	D flip-flop with set and reset	Yes	Yes	—
	Divide by 2 flip-flop w/jam load and reset	—	Yes	Yes
	Divide by 2 flip-flop w/reset	—	Yes	—
	D flip-flop w/jam load and reset	—	Yes	Yes
	D flip-flop shift register with reset	—	Yes	—
JK Flip-Flops	JK flip-flop with reset			
	JK flip-flop with			
T Flip-Flops				
			res	—
		—	Yes	Yes
		—	Yes	—
		—	Yes	Yes
	counter w/jam load and reset	—	Yes	Yes
	Up/down counter with reset	—	Yes	—
	Up/down counter w/jam load and reset	—	Yes	Yes

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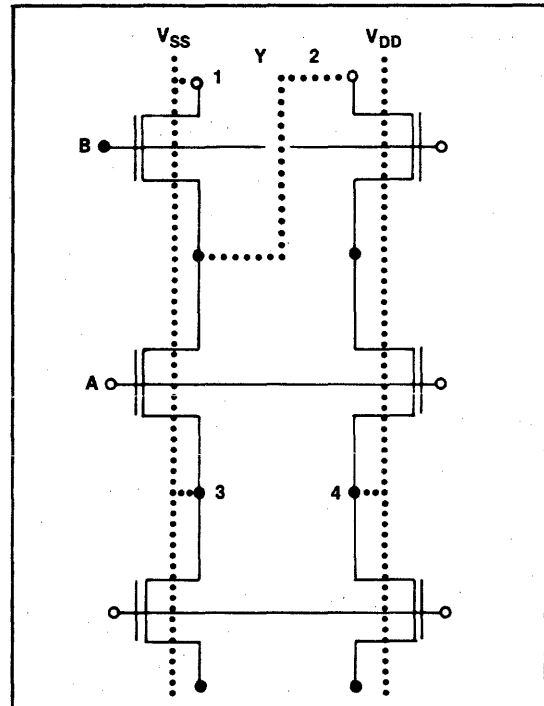


Figure 5. Schematic Diagram for the 2-Input NOR

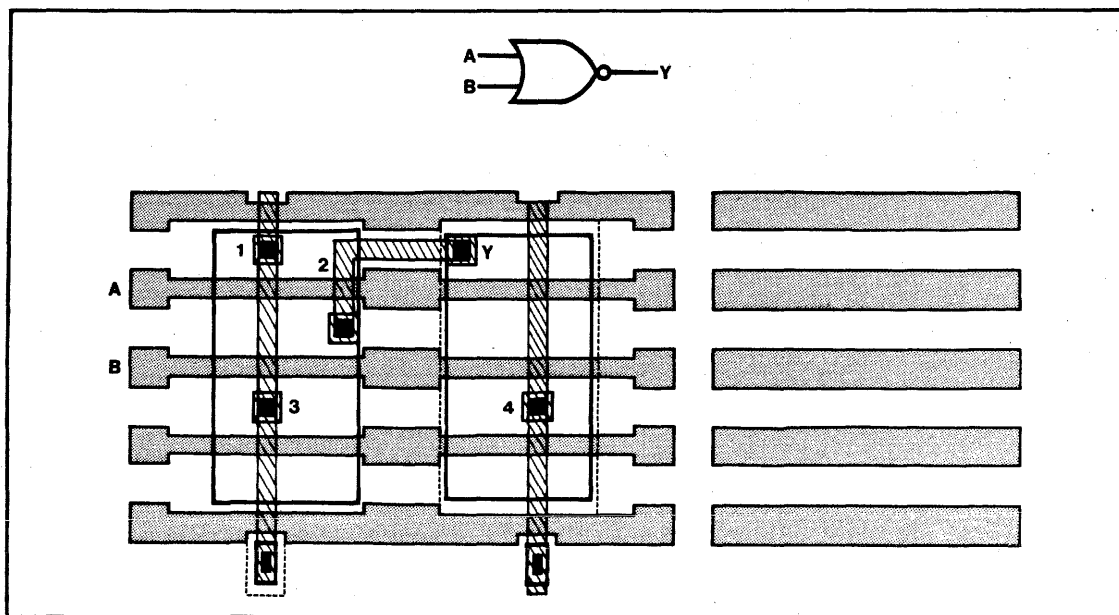


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Multiplexer	2 to 1 multiplexer			
Schmitt Trigger	Schmitt Trigger			
Transmission Gate	Buffered transmission gate Unbuffered transmission gate			
Tristate Control	Tristate control			
IGC10000 Sequential Macros		Set	Reset	Jam Load
D Flip-Flops	D flip-flop with reset	—	Yes	—
	D flip-flop with set	Yes	—	—
	D flip-flop with set and reset	Yes	Yes	—
	Divide by 2 flip-flop w/jam load and reset	—	Yes	Yes
	Divide by 2 flip-flop w/reset	—	Yes	—
	D flip-flop w/jam load and reset	—	Yes	Yes
	D flip-flop shift register with reset	—	Yes	—
	JK Flip-Flops	JK flip-flop with reset	—	Yes
JK flip-flop with set		Yes	—	—
T Flip-Flops	T flip-flop with reset	—	Yes	—
Latches	D latch	—	—	—
	D latch w/single input control	—	—	—
	D latch with reset	—	Yes	—
	D latch w/reset and single input control	—	Yes	—
	D latch w/transmission gate on output	—	—	—
Counters	Down counter with reset	—	Yes	—
	Down counter w/jam load and reset	—	Yes	Yes
	Up counter with reset	—	Yes	—
	Up counter w/jam load and reset	—	Yes	Yes
	Up/down counter with reset	—	Yes	—
	Up/down counter w/jam load and reset	—	Yes	Yes

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Table 2. IGC10000 MACRO Library (continued)

IGC10000 Digital I/O Cell Macros	
Type	Description
Bidirectional	Tristate output/unbuffered input Tristate output/inverting TTL input buffer
Feedthrough	Input feedthrough
Internal Buffer	Inverting internal buffer Internal tristate buffer
Input Buffer	Non-inverting TTL input buffer Non-inverting CMOS input buffer with pull-up options
Output Buffer	Open drain output buffer Non-inverting TTL output buffer Inverting TTL output buffer Tristate output buffer Symmetrical drive output buffer
IGC10000 Analog I/O Cell Macros	
<p>The list below represents a sample of custom macros developed for specific analog applications. Consult your Intersil Representative for suitability to your design.</p> <ul style="list-style-type: none"> <li>Analog Transmission Gate</li> <li>Auto Null Comparator</li> <li>Comparator</li> <li>Compensated Op Amp</li> <li>Crystal Oscillator</li> <li>Current Multiplier</li> <li>Current Reference</li> <li>Op Amp</li> <li>Power-on Reset</li> <li>RC Oscillator</li> <li>Schmitt Trigger</li> </ul>	

**COMPUTER AIDED DESIGN SOFTWARE TOOLS**

The IGC10000 family is supported by a proprietary computer aided design (CAD) system developed at the General Electric Microelectronics Center. The system provides CAD tools for logic simulation, accurate prediction of circuit speed performance, automated design of interconnect circuitry, electrical and design rule checking, post-layout simulation using RC delays extracted from the layout, and automatic conversion of simulation test pattern files into tester format.

The CAD tools are integrated under a supervisory program called the CADEXEC (for CAD Executive) that runs on a Digital Equipment Corporation VAX com-

puter. Once the user has entered the circuit's interconnect information into the computer, this information is converted into a common database accessed by all other parts of the software through the CADEXEC.

**Logic Simulation:** Users have access to the TEGAS logic simulator. For pre-layout logic (functional) verification, customers may perform TEGAS simulation using our Unit Delay Macro Library database; for design verification (pre-layout timing analysis), calculated delays based on fanout are used in conjunction with Best, Typical, and Worst Case libraries, whose parameters are described in Table 3.

**Table 3. Best, Typical, and Worst Case Parameters**

Parameter	Best	Typical	Worst
Voltage (V)	5.5	5.0	4.5
Temperature (°C)	0	27	70
Process	Best	Typical	Worst

**Routing:** The SILICA layout system uses a proprietary automatic router developed at the General Electric Microelectronics Center. The SILICA router has consistently performed with higher completion rate and lower CPU time than other commercially available routers; in addition, the router improves the overall performance of the circuit by selecting paths that produce the smallest delay. Like many routers, the SILICA router has a Critical Net feature that minimizes polysilicon and total net length by routing the critical net first. Unique to the SILICA router is the Super Critical Net feature, which prohibits the use of polysilicon gates in a specified net.

**Electrical and Design Rule Checking:** SILICA DRC (Design Rule Checker) performs electrical and design rule checking in minutes instead of hours and extracts geometric data from the layout for input into the RC delay extraction software.

**Manual editing:** In rare instances manual layout editing may be done on one of our CALMA workstations. CALMA output is fed into SILICA DRC for convenient verification of electrical integrity.

**Post-layout simulation:** A specialized circuit simulator has been developed at the General Electric Microelectronics Center to compute the delays of the RC-interconnect nets from the topology of the network after performing layout. The RC Delay extraction software uses a full transient analysis for each net; delays are based on resistance as well as capacitance of the interconnection nets. The software calculates delays as a function both of load switching voltage and driver output impedance and handles loops, bidirectional drivers, and multiple drivers on the net.

After the RC delay information is extracted, the CADEXEC system inserts the delays into the network database for post-layout TEGAS simulation and critical path analysis, thus providing an additional opportunity for refining the layout prior to PG tape generation.

**Tester Tape Generation:** Test program conversion software automatically translates the customer's final TEGAS simulation output file into a test vector pattern file to be used in testing the finished device.

**PACKAGING**

Five types of packages are available for the IGC10000 gate arrays. Dual inline packages are available in plastic (Plastic DIP), ceramic (CerDIP) and multilayer ceramic (Side Brazed DIP); leadless chip carriers and pin grid arrays are provided in multilayer ceramic. Table 4 presents recommended package types for each pin count and array size.

**Table 4. Recommended Package Types**

Number of Pins	Plastic DIP	CerDIP	Side Brazed DIP	Leadless Chip Carrier	Pin Grid Array
8	408		408		
14	408	408	408		
16	408	408	408		
18	408	408	408		
20		408	408		
24	408	408	408		
	756	756	756		
	1500	1500	1500		
28	408	408	408		
	756	756	756		
	1500	1500	1500		
40	408	408	408		
	756	756	756		
	1500	1500	1500		
44				756	
				1500	
48			756		
			1500		
52				1500	
68				1500	1500

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**DEVELOPMENT**

An overview of the gate array development process is shown in the flow chart of Figure 6. During Phase 1 (Design Translation), most of the responsibility lies with the customer; during Phase 3 (Fabrication), with Intersil. In Phase 2 (Design Implementation), most of the activities are performed by Intersil, but require

customer interaction and approval. Figure 6 delineates the responsibilities of the customer and of Intersil. For more information, contact either your local Intersil representative, or Semicustom Marketing at the General Electric Microelectronics Center, Research Triangle Park, NC, telephone 919-549-3607.

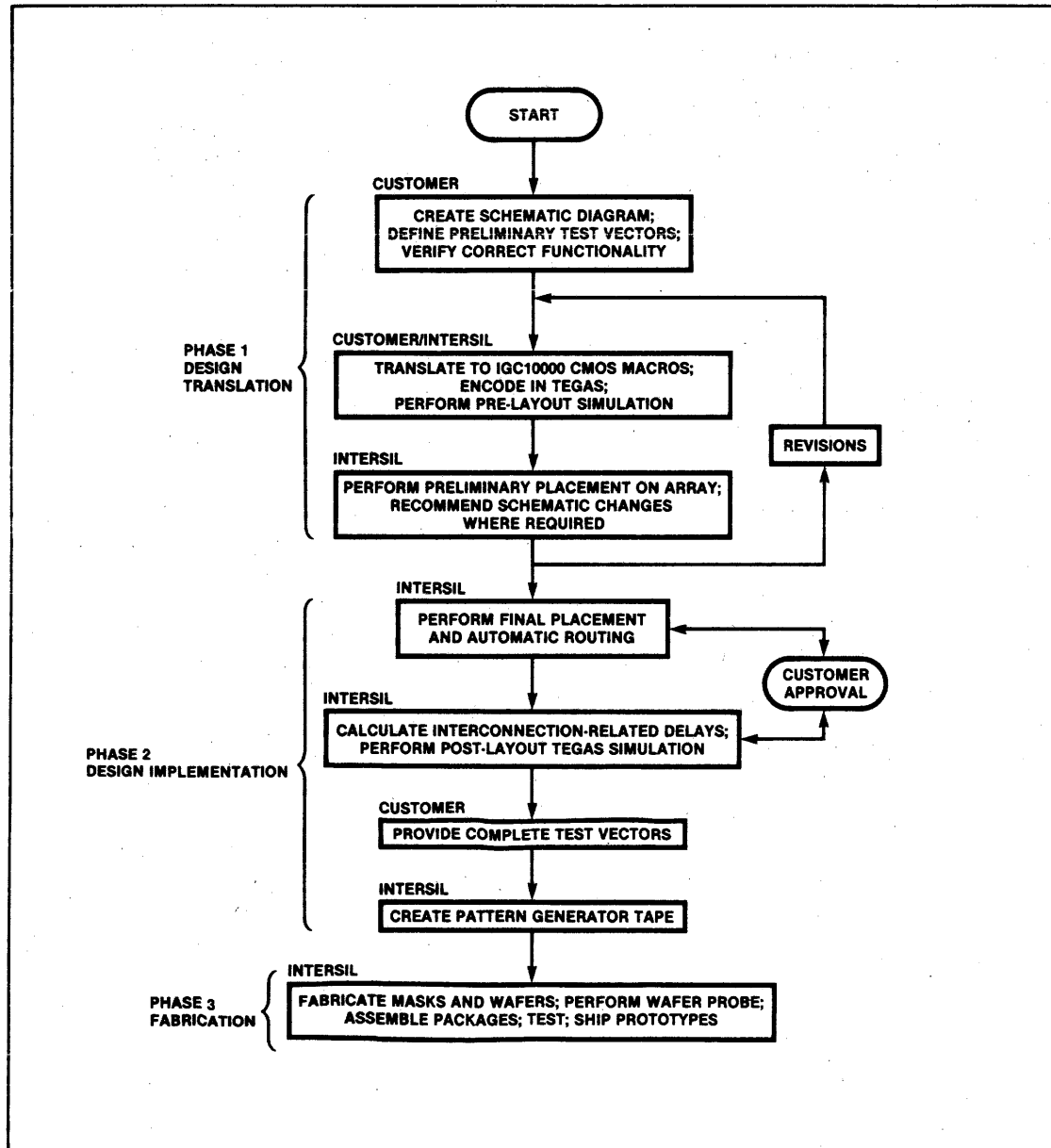


Figure 6. Simplified Flowchart for Gate Array Development



**OPERATING CHARACTERISTICS<sup>1</sup>**

Absolute Maximum Ratings<sup>2</sup> (Referenced to V<sub>SS</sub>)

Parameter	Symbol	Limits	Units
DC Supply Voltage	V <sub>DD</sub>	- 0.5 to + 10.0	V
Input Voltage	V <sub>I</sub>	- 0.5 to V <sub>DD</sub> + 0.5	V
DC Input Current	I <sub>I</sub>	± 10	mA
Operating Ambient Temperature Range	T <sub>A</sub>	- 55 to + 125	°C
Storage Temperature Range (Ceramic)	T <sub>STG</sub>	- 65 to + 150	°C
Storage Temperature Range (Plastic)	T <sub>STG</sub>	- 40 to + 125	°C

**NOTE 1:** Stress ratings only. Functional operation of the device at these or any conditions beyond those indicated as Recommended Operating Conditions is not implied.

**NOTE 2:** Stresses above those listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Recommended Operating Conditions**

Parameter	Symbol	Limits	Units
DC Supply Voltage	V <sub>DD</sub>	3.3 ± 0.3V to 9.0 ± 0.9V	V
Typical Operating Frequency	f <sub>CK</sub>	8.0	MHz
Operating Ambient Temperature Range <sup>1</sup>	T <sub>A</sub>	- 55 to + 125	°C

**NOTE 1:** IGC10000 gate array macros are currently characterized between 0 and 70°C.

**AC CHARACTERISTICS**

Specified for nominal processing = 5V, 27°C.  
 Calculated for a fanout of 1.

	Parameter	Typical Delay (ns)
<u>Array Cell Macros</u>		
2-input NAND	D to Output	6
2-input NOR	D to Output	6
4-input NAND	D to Output	8
4-input NOR	D to Output	18
1X inverter	D to Output	5
4X inverter	D to Output	4
2-1 AND-OR invert	D to Output	9
D flip-flop with reset	CK to Output	9
Schmitt trigger	Input to Output	18
Up counter with reset	CK to Output	11
<u>I/O Cell Macros</u>		
Input feedthrough	Pad to Output	1
Non-inverting Input Buffer	Pad to Output	9
Non-inverting Output Buffer	D to Pad	10 (15 pF) 19 (50 pF)

**DC CHARACTERISTICS**

$V_{DD} = 5V \pm 10\%$

Symbol	Parameter	Condition	Limits <sup>1</sup>							
			0°C		25°C			70°C		
			Min.	Max.	Min.	Typ.	Max.	Min.	Max.	Units
$I_{DD}^2$	Quiescent Device Current	$V_I = V_{DD}$ or $V_{SS}$				0.3			100	$\mu A$
$V_{OL}$	Low Level Output Voltage	$ I_O  \leq 1\mu A$		0.05				0.05	0.05	V
$V_{OH}$	High Level Output Voltage	$ I_O  \leq 1\mu A$	$V_{DD} - 0.05$		$V_{DD} - 0.05$			$V_{DD} - 0.05$		V
$V_{IL}$	Low Level Input Voltage	CMOS I/O Macro		1.5				1.5	1.5	V
$V_{IH}$	High Level Input Voltage	CMOS I/O Macro	3.5		3.5			3.5		V
$V_{IL}$	Low Level Input Voltage	TTL I/O Macro		0.8				0.8	0.8	V
$V_{IH}$	High Level Input Voltage	TTL I/O Macro	2.0		2.0			2.0		V
$I_{OL}^3$	Output Low <sup>4</sup> (Sink Current)	$V_O = 0.4V$	1.8		1.8	3.6		1.6		mA
		$V_O = 2.5V$	3.8		3.8	7.6		3.4		mA
$I_{OH}^3$	Output High (Source Current)	$V_O = 4.6V$	0.3		0.3	0.6		0.25		mA
		$V_O = 2.5V$	1.8		1.8	3.6		1.6		mA
$I_{IN}$	Input Leakage Current	$V_{IN} = 0$ or $V_{DD}$		$\pm 0.1$		$\pm .001$	$\pm 0.1$		$\pm 1.0$	$\mu A$
$I_{OZ}$	Tristate Output Leakage Current	$V_O = 0$ or $V_{DD}$		$\pm 1.0$		$\pm .001$	$\pm 1.0$		$\pm 10$	$\mu A$
$C_{IN}$	Input Capacitance	Any Input				5.0				pF

**NOTES:**

- IGC10000 gate arrays are designed to perform under conditions up to 125°C. Limits reflect temperature range at which the macro library is characterized.
- Any internal oscillators disabled.
- Results depend on specific output macro used.
- There may be limitations on maximum current when many outputs are simultaneously low.

# ABBREVIATIONS OF COMPANY NAMES

**Action Ins** Action Instruments  
**AD** Analog Devices  
**ADT** Advanced Digital Technology  
**Advent** Advent Products, Inc.  
**Alphatron** Alphatron  
**AMA** American Automation  
**AMD** Advanced Micro Devices  
**AMI** American Microsystems, Inc.  
**Amperex** Amperex Electronic Corp.  
**Analogic** Analogic  
**Analog Sys** Analog Systems  
**APC** Applied Micro Circuits  
**Apex** Apex Microtechnology  
**APM** Applied Microsystems Corp.  
**Appl Sys** Applied Systems Corp.  
**APT** Applied Microtechnology  
**Aptek** Aptek Microsystems  
**Array Tech** Array Technology  
**AWI** AWI Electronics

**Barvon** Barvon Research  
**Bedford** Bedford Computer Systems Inc.  
**Burr-Brown** Burr-Brown

**CAE** Computer Aided Engineering  
**Cal Devices** California Devices  
**Cermetek** Cermetek  
**CGRS** CGRS Microtech Inc.  
**Cherry** Cherry Semiconductor  
**CIC** Custom Integrated Circuits  
**CirTech** Circuit Technology  
**Citel** Citel  
**Comlinear** Comlinear Corporation  
**CMA** Custom MOS Arrays  
**Comark** Comark Corp.  
**Comdial** Comdial Semiconductor  
**Comp Auto** Computer Automation  
**Compas** Compas Microsystems  
**Cont Logic** Control Logic Inc.  
**Control Sys** Control Systems Microsystems Div.  
**CreMicro** Creative Micro Systems  
**Cromemco** Cromemco, Inc.  
**Cubit** Cubit Inc.  
**Curtis** Curtis Electro Devices, Inc.  
**Cybernetic** Cybernetic Micro Systems  
**Cybersys** Cybersystems  
**Cybertek** Cybertek Inc.

**Data General** Data General  
**Data I/O** Data I/O  
**Data Trans** Data Translation  
**Datel** Datel  
**Datronic** Datronic Corporation  
**DDC** Data Devices Corporation  
**DEC** Digital Equipment Corporation  
**Die-Tech** Die-Tech  
**Digelec** Digelec Corp.  
**Digitek** Digitek, Inc.  
**Dionics** Dionics Inc.  
**Dist Comp** Distributed Computer Systems  
**Divers Tech** Diversified Technology

**E-HI** E-H International, Inc.  
**EDI** Electronic Designs Inc.  
**Elind** Elind Electronica Industriale  
**EMM-SESCO** EEM-SESCO  
**Emulogic** Emulogic Inc.  
**ETI Micro** ETI Micro  
**Exar** Exar Integrated Systems  
**Exel** Exel Microelectronics

**Fairchild** Fairchild  
**Ferranti** Ferranti Electric  
**Force** Force Computers  
**Fujitsu A** Fujitsu America  
**Fujitsu** Fujitsu Microelectronics, Inc.

**GI** General Instrument  
**GTE Micro** GTE Microcircuits

**Harris** Harris Semiconductor  
**Heurikon** Heurikon Corp.  
**Hilevel** Hilevel Technology, Inc.  
**Hitachi** Hitachi America, Ltd.  
**Holt** Holt Inc.  
**HP** Hewlett-Packard  
**Hughes** Hughes Aircraft, Solid State Products

**Hybrid Sys** Hybrid Systems  
**HyComp** HyComp

**ICC** International Cybernetics  
**IDT** Integrated Device Technology  
**IMI** International Microcircuits, Inc.  
**IMP** International

**IMS** Microelectronic Products  
**Infosphere** Industrial MicroSystems Inc.  
**Inmos** Infosphere  
**IntCirEng** Integrated Circuit Engineering  
**IntCirSys** Integrated Circuit Systems  
**IntCompSys** Integrated Computer Systems  
**Int Tech** Integrated Technology Corp.  
**Intech** Intech Microcircuits Div.  
**Intel** Intel  
**Interdesign** Interdesign  
**Intersil** Intersil  
**Intronics** Intronics  
**ITT** ITT Semiconductors

**Kinetic Sys** Kinetic Systems  
**Kontron** Kontron Electronics

**Lambda** Lambda Semiconductor  
**Linear Tech** Linear Technology  
**LSI Comp** LSI Computer Systems  
**LSI Logic** LSI Logic Corporation

**Master Logic** Master Logic Corporation  
**Matrix** Matrix Corp.  
**Matrox** Matrox Electronic Systems  
**Maxim** Maxim Integrated Products  
**MCC** Micro-Computer Control  
**MCE** MCE Electronics  
**Micrel** Micrel  
**Micro Innov** Micro Innovators  
**Micropac** Micropac Industries  
**Micro Net** Micro Networks  
**Micro Pwr** Micro Power Systems  
**Micro Sci** Micro Sciences Corp.  
**Micro Tech** Microcircuits Technology  
**Micro-Link** Micro-Link Corporation  
**Micron** Micron Technology  
**MilerTron** MilerTronics  
**Miller** Miller Technology  
**Mitel** Mitel Semiconductor  
**Mitsubishi** Mitsubishi Electronics  
**MMI** Monolithic Memories, Inc.

**Mostek** Monolithic Systems Corp.  
**Motorola** Mostek  
**MRC** Motorola Semiconductor  
**Murray** MRC Systems  
**Monosil** Murray Consulting

**National** National Semiconductor  
**NCM** NCM Corp.  
**NCR** NCR Corp., Microelectronics Division  
**NEC** NEC Electronics  
**Nitron** Nitron

**OAE** Oliver Advanced Engineering  
**Octagon** Octagon Systems Corp.  
**OEI** Optical Electronics Inc.  
**Ohio Sci** Ohio Scientific  
**OKI** OKI Semiconductor  
**Omnibyte** Omnibyte Corp.  
**Onset** Onset Computer Corp.

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**Pico Design** Pico Design  
**Polycore** Polycore Electronics  
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**PMI** Precision Monolithics, Inc.  
**PragDes** Pragmatic Design Inc.  
**Pro-Log** Pro-Log Corp.

**Quay** Quay Corp.

**Raytheon** Raytheon Semiconductor  
**RCA** RCA Solid State Division  
**RCI Data** RCI Data  
**RELMS** Relational Memory Systems  
**Reticon** Reticon  
**RIFA** RIFA  
**Rockwell** Rockwell, Microelectronic Devices  
**RTC** Riehl Time Corporation

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**SBE** SBE, Inc.  
**SEEQ** SEEQ Technology, Inc.  
**SPI** Semi Processes Inc.  
**Siemens** Siemens  
**Si-Fab** Si-Fab  
**Signetics** Signetics  
**SGS** SGS Semiconductor  
**Sharp** Sharp  
**Silicon G** Silicon General  
**Siliconix** Siliconix  
**Silicon Sys** Silicon Systems Inc.  
**Siltronics** Siltronics  
**SMC** Standard Microsystems Corp.  
**S MOS** S MOS Systems  
**Solarise** Solarise Enterprises  
**Solitron** Solitron Devices  
**Sprague** Sprague Electric Company  
**SSM** Solid State Micro Technology for Music

**SSS** Solid State Scientific  
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**Zytrex** Zytrex Corp.



**If you're thinking of custom,  
watch where you're going**



**WHEN PICKING A CUSTOM LSI MANUFACTURER, THE FIRST THING TO LOOK AT IS HOW GOOD THEY ARE. THE SECOND IS WHERE THEY'RE LOCATED.**

If you're even thinking about a custom circuit, you're aware of the benefits it will give you: it will, over the long run, cost less to produce than any form of semi-custom; by incorporating the maximum number of functions, it will give you the most savings in board space, shipping, inventory control, inspection, manufacturing operations, and reliability. A properly designed custom chip will greatly increase your product's versatility, and give you an enormous marketing advantage.

**But how can you be sure you'll actually get all these benefits?**



**ARE THEY CUSTOM SPECIALISTS?**

Even with all the computerized assistance available today, it takes a trained specialist to give you a custom chip that lives up to your expectations. It takes years of experience to design the chip that will work most effectively with your system; to know how many functions can safely be included, i.e., how much can be included without risk of the chip's not working. *LSI Computer*

*Systems, Inc., has been creating custom MOS/LSI circuits for over 14 years. That's a long, long time, in this business.*



**ARE THEY EASY TO WORK WITH?**

The second way to make sure you get the most out of your custom chip is to work very closely with the manufacturer's engineers. The more active the interplay between you (or your engineers) and the manufacturer's designers, the more your chip will do for you (and, we might add, the faster the design will be completed).

And it's a lot easier to work closely with somebody when you can step in your car and be at his place in an hour or so. Or vice versa.

Well, if you're anywhere within 50 miles of New York, we — LSI Computer Systems, Inc. — are just such an easy drive away. We're right smack in the middle of Long Island.

Of course, being "easy to work with" is not solely a function of geographical location. It's also a function of accessibility, and *that's* largely a function of size. Many of our customers who have also worked with big LSI manufacturers tell us that our engineers are far more accessible than the big outfits' engineers are. We'd like to be big someday, too, but accessible is one thing we'll always be. It's one of the things that made us as big as we are.

**Call Ron Colino.** He's our VP/Marketing, and he can tell you a lot more reasons why you should come with us if you're thinking of custom. He can also tell you a lot of reasons why you *should* be thinking of custom if you're not.

**LSI**  
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Please see IC Master, DIGITAL SECTION, page 803 for our line of Standard micro-circuits.

**Features**

- Silicon-gate 3.5-micron (drawn) HCMOS technology
- Single-layer metal interconnection
- HTTL and LSTTL speeds—5 ns through a 2-input NAND gate and interconnection,  $T_A = 25$  degrees C, fanout = 2,  $V_{DD} = 5V$
- Optimal block structure of 2N and 2P transistors
- Complexities ranging from 272 to 2550 blocks
- Pin counts ranging up to 104
- Fully supported by LDS™ (LSI Design System)
- Extensive macrocell and macrofunction libraries
- All non-power pads configurable as inputs, outputs or bidirectional
- TTL/CMOS I/O compatibility
- Configurable output drive up to 4.8mA
- All inputs and outputs protected from over-voltage and latch-up
- Full Military capability
- Ceramic and plastic packages
- LL3110Q evaluation device available

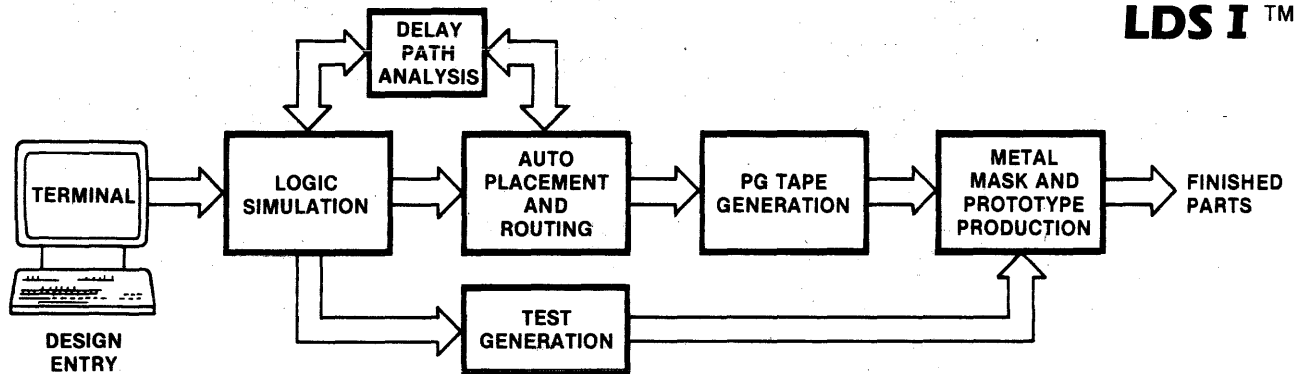
**Features**

- Silicon-gate 3-micron (drawn) HCMOS technology
- Schottky TTL speeds - 2.5ns through 2-input NAND gate and interconnection,  $T_A = 25^\circ C$ , fanout = 2,  $V_{DD} = 5V$
- Optimal block structure of 2N and 2P transistors
- Complexities ranging from 880 to 6000 blocks
- Pin counts ranging up to 180
- Fully supported by LDS™ (LSI Design System)
- Extensive macrocell and macrofunction libraries
- All non-power pads configurable as inputs, outputs or bidirectional
- TTL/CMOS I/O compatibility
- Configurable output drive up to 6.0mA
- All inputs and outputs protected from over-voltage and latch-up
- Full military capability
- Ceramic and plastic packages
- Alternately-sourced
- LL5220Q evaluation device available

**Features**

- Silicon-gate 2.0-micron (drawn) HCMOS technology
- Faster than Schottky TTL speeds - 1.4ns through 2-input NAND gate and interconnection,  $T_A = 25^\circ C$ , fanout = 2,  $V_{DD} = 5V$
- Optimal block structure of 2N and 2P transistors
- Complexities ranging from 968 to 10,013 blocks
- Pin counts ranging up to 180
- Fully supported by LDS™ (LSI Design System)
- Extensive macrocell and macrofunction libraries
- All non-power pads configurable as inputs, outputs or bidirectional
- TTL/CMOS I/O compatibility
- Configurable output drive up to 6.0mA
- All inputs and outputs protected from over-voltage and latch-up
- Full military capability
- Ceramic and plastic packages
- Alternately-sourced
- LL7220Q evaluation device available

**LSI Development System  
LDS I™**



**Development Support and Interface**

The LDS I™ (LSI Development System I) is used for logic simulation, design verification, test pattern generation, layout and mask design. The system may be accessed using a variety of remote terminals, including the Tektronix 4014 and 4112 vector-display terminals. The desired logic function may be entered alpha numerically or schematically into the system and the design electronically breadboarded using the LDS I™ logic simulator. The simulated input patterns may be checked for completeness, and other tests added (automatically if desired) until the final test pattern is completed. The personality masks may then be designed remotely by the customer using the LDS I™ interactive/automatic placement and routing program. Alternately, LSI LOGIC CORPORATION will complete the metal mask design. For more details see the LDS I™ Product Description available from LSI LOGIC CORPORATION.

For new designs, the LDS I™ Development System may be used to verify the design, generate test patterns and (optionally) design the metal masks. Four possible interfaces to LDS I™ are offered.

1. Use the complete LSI LOGIC CORPORATION system in Milpitas, California.

2. Install a remote terminal in your facility connected to LSI LOGIC CORPORATION mainframe. The Tektronix 4014 and 4112 are both supported, with the 4112 providing significantly higher performance. In either case a Tektronix 4662 11-inch×17-inch plotter may be used for graphics hard copy.
3. Install a turnkey LDS I™ system in your facility.
4. LSI Logic can specify design rules and provide macro layouts so that a customer's own CAD system may be used.

The following material is required by LSI LOGIC CORPORATION to alternate source an existing CDI-HC or AMI-UA array part.

- Calma data base or PG tape of the metal mask
- Sentry test tape
- AC test specifications
- Package and marking specifications
- Bonding diagram
- Four devices meeting all required specifications

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# MICREL SEMICONDUCTOR SERVICES

- Alternate/Discontinued Device Source
- Custom ICs
- Wafer Processing
- Wafer/Packaged IC Testing

## ALTERNATE/DISCONTINUED DEVICE SOURCE\*

SY2827	MIC2827	2048 Dynamic Shift Register
SY2825	MIC2825	Dual 1024 Bit Dynamic Shift Register
SY1404	MIC1404	1024 Bit Dynamic Shift Register

\* Call Micrel if you need help in Device Sourcing.

## WAFER PROCESSING

**Technology:** PMOS, CMOS, NMOS Silicon or Metal Gate, Bipolar Linear. Minimum feature size is 4 micron and Isoplaner type processing is available for Silicon Gate MOS. Epi thickness 6-12 micron with Arsenic/antimony buried layer. 3 or 4 inch wafers.

**Clean Room:** 5000 ft<sup>2</sup> of class 100 clean room.

**Engineering Run:** 25 wafer starts with 10 guaranteed to meet your Process parameters. Delivery is 5 weeks with plates.

**Production Run:** 25 wafer minimum with no minimum annual volume. 10,000 wafer starts per month capacity. Delivery is 4 weeks, expedited to 10 days.

**Customer Supplies:** Tooling, design criteria and test software. We'll honor non-disclosure agreements.

**Ion Implant:** Varian DF4U/Veeco 2100

## WAFER/PACKAGED IC TESTING:

**Wafer Test:** Map and test to process parameters.

**Packaged IC Test:** Mil STD 883 B Group A, using high/low temperature handlers.

**Wafer Probing/Packaged IC:** 3 Sentry VIIs or Series 20 VLSI Testers with handlers.

**SEM Analysis:** Mil STD 883B using ISI 5 x 30, 5-30KEV.

**Screening:** Mil STD 883B class B or customer requirement.



# GATE ARRAY MYTHS vs.

## MYTHS ..... vs. .... REAL CUSTOM

### MYTH 1 — User Friendly

"My logic diagram will always work when put in a gate array. A myth!"

### MYTH 2 — No System Timing Problems

"Circuit will never have any race conditions. A myth!"

### MYTH 3 — Development Times Are Never Extended

"Chip will always work the first time, and never use more than one mask. A myth!"

### MYTH 4 — No Customer Risk

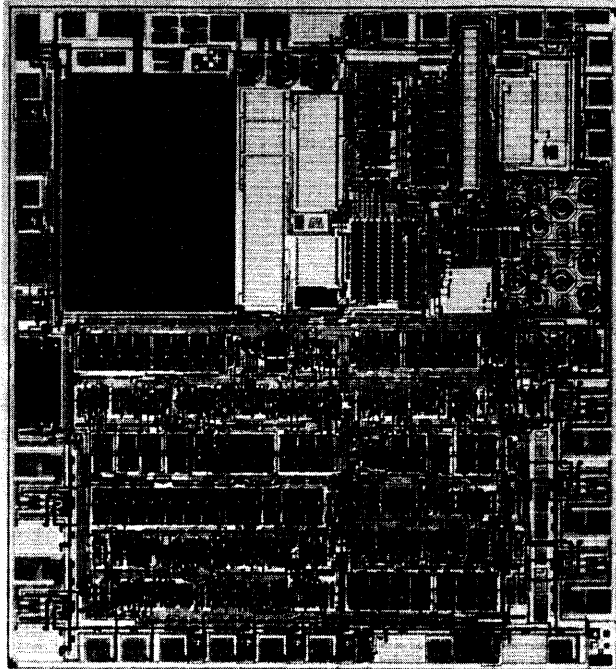
"Development cost is low. A myth!" (Actually you pay double if chip doesn't work the first time.)

### MYTH 5 — Testing Is Simple

"A myth!" (Actually testing is more difficult than the design.)

### MYTH 6 — You're Never Switched from Gate Array to Full Custom

"The biggest myth of all."



MPS Real Custom Chip (\*HALLMOS circuit)

### MYTH $\infty$

"Quick. Take your logic diagram to any gate array company and in a few days/weeks — and almost like magic — tested production parts will appear, at a very low price. Now you can beat out your competitors."

### FACTS

At Micro Power Systems, prospects are taken through the *Gate Array vs. Real Custom* analysis. Facts not myths determine the next step: (1) MPS' Real Custom, (2) MPS' Full Custom, or (3) other's gate arrays.

### REAL CUSTOM

When you work in the real world of today — you need MPS' Real Custom to achieve your design and marketing goals. Over 65% of Micro Power's prospects fall into the Real Custom category. Here MPS uses an advanced, low-cost standard cell

approach (over 2000 cells) for 80% of the design and 20% new design to optimize the interconnects and provide system enhancements. Quick turn-around and lower production costs are achieved by a proprietary interconnect system that provides a fast layout technique and optimizes silicon utilization.

### FULL CUSTOM

Another 10% of our prospects take advantage of MPS' full custom. This is a 9 to 12-month joint effort where advanced circuit design, system definitions, and custom processing provide a quantum jump over the competition.

\*HALLMOS is a major breakthrough technology that combines precision linear bipolar, high accuracy thin film resistors, and high density CMOS all on the same monolithic integrated circuit. This process has a big potential in systems that utilize both analog and digital functions.




## GATE ARRAYS

About 20 to 25% of our prospects require less than 500 parts per month. These prospects need a gate array company. We suggest they look for solid companies with their own foundries, testing, engineering support, and with realistic pricing and scheduling.

### 1 REAL CUSTOM vs. 1 GATE ARRAY

Compare MPS' Real Custom against one super gate array (see chart). Even comparing one super gate array shows major advantages for the Real Custom IC. Not only do you have significant parts and testing advantages — you can't use any precision analog in the super gate array. Is there a big difference between gate array and Real Custom delivery schedules? No! First, the gate array companies don't quote on a complete schedule. They use the time from when your circuit design is accepted (i.e., after you change your circuit to meet their technology limitations). Second, they quote to the most optimistic delivery date of the first prototype. You should prepare a realistic schedule with all the required events (yours and the vendors) from concept to delivery of production quantities. Then compare it to a Real Custom schedule and you'll find that they provide very little time savings, if any.

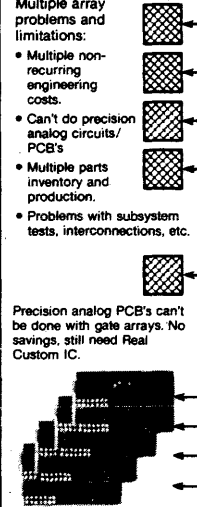
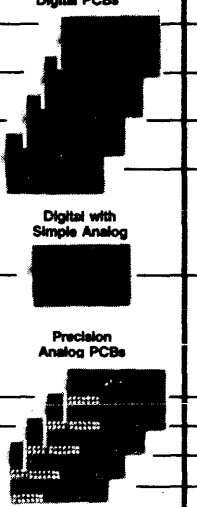

### One Super Gate Array vs. One Real Custom IC

SUPER GATE ARRAY	SYSTEM	REAL CUSTOM
<p>One large super gate array (2000 to 5K gates), plus many peripheral ICs, resistors, etc. Precision analog PCB required (no precision analog on gate array).</p> 		<p>One Real Custom IC, plus a few discrete components. No peripheral costs. Includes precision analog.</p> 

### 1 REAL CUSTOM vs. MULTIPLE ARRAYS

Look at the significant difference between using multiple gate arrays and one Real Custom IC in a digital/analog subsystem (see chart). Don't forget to add in your multiple gate array costs and costs for system definition, interconnection, and testing. With arrays, extensive interconnections remain at the system/board level and system testing becomes very difficult. Implementation of precision analog circuits are not even considered practical on arrays. Real Custom has inherently lower system development costs and lower production costs over solutions requiring multiple gate arrays.

### Standard Gate Arrays vs. One Real Custom IC

GATE ARRAYS	SUBSYSTEM	REAL CUSTOM
<p>Multiple array problems and limitations:</p> <ul style="list-style-type: none"> <li>Multiple non-recurring engineering costs.</li> <li>Can't do precision analog circuits/PCBs</li> <li>Multiple parts inventory and production.</li> <li>Problems with subsystem tests, interconnections, etc.</li> </ul> <p>Precision analog PCB's can't be done with gate arrays. No savings, still need Real Custom IC.</p> 	<p>Digital PCBs</p> <p>Digital with Simple Analog</p> <p>Precision Analog PCBs</p> 	<p>One integrated subsystem with:</p> <ul style="list-style-type: none"> <li>Subsystems tested.</li> <li>90% interconnections.</li> <li>Precision analog included.</li> <li>One chip inventory and production.</li> <li>One non-recurring engineering cost.</li> </ul> <p>One Real Custom IC</p> 

### CUSTOM CHIPS DELIVERED IN 1971

MPS has delivered full custom and Real Custom for over 13 years. We have advanced processes and design capabilities immediately available. For example, we routinely combine low noise bipolar transistors, CMOS logic, and thin film resistors on the same monolithic circuit. All these advanced technologies are currently in volume production.

### RUN-AROUND

If you're in management and your personnel have been giving you the run-around about why your gate array project has gone through several re-designs and why the schedule has slipped — you'd better check out what the new projected production costs (piece price) are going to be. It may no longer fit into your marketing plan.

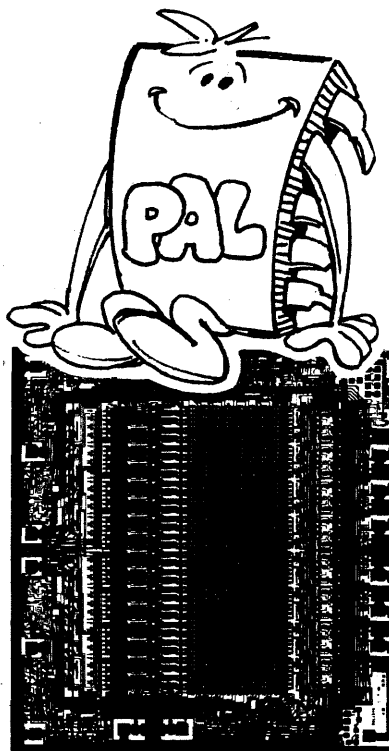
### MYTH PERSISTS

We can't stop the gate array myth, but if you're serious about your company's future and being second best is not acceptable, then send for our *MPS Real Custom Plan*, or call Bob Sabo (ext. 210). A gate array shakeout is happening now. Do you want to deal in myths or facts?

**MICRO  
POWER  
SYSTEMS**

3100 Alfred Street  
Santa Clara, CA 95050  
(408) 727-5350

*Micro Power Systems is the Leader in Refractory Gate MOS Technology*



## The PAL<sup>®</sup> Concept

Monolithic Memories' family of PAL devices gives designers a powerful tool with unique capabilities for use in new and existing logic designs. The PAL saves time and money by solving many of the system partitioning and interface problems brought about by increases in semiconductor device technology.

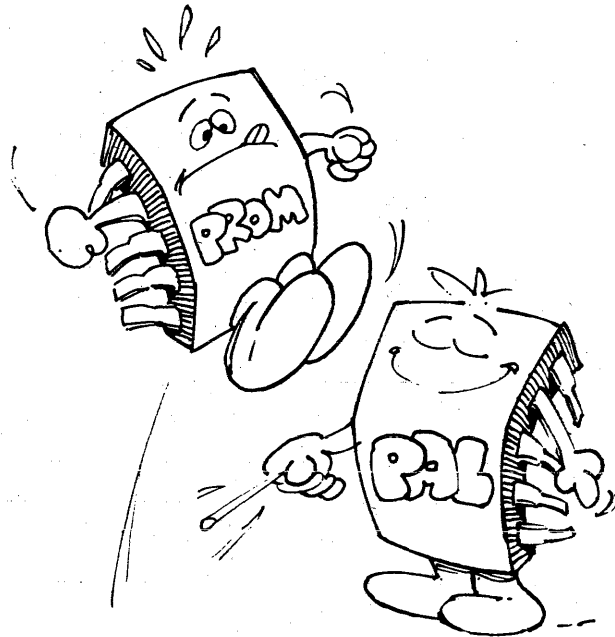
Rapid advances in large scale integration technology have led to larger and larger standard logic functions; single I.C.s now perform functions that formerly required complete circuit cards. While LSI offers many advantages, advances have been made at the expense of device flexibility. Most LSI devices still require large numbers of SSI/MSI devices for interfacing with user systems. Designers are still forced to turn to random logic for many applications.

The designer is confronted with another problem when a low to medium complexity product is designed. Often the function is well defined and could derive significant benefits from fabrication as an integrated circuit. However, the design cycle for a custom circuit is long and the costs can be very high. This makes the risk significant enough to deter most users. The technology to support maximum flexibility combined with fast turn around on custom logic has simply not been available. Monolithic Memories offers the programmable solution.

The PAL family offers a fresh approach to using fuse programmable logic. PAL circuits are a conceptually unified group of devices which combine programmable flexibility with high speed and an extensive selection of interface options. PAL devices can lower inventory, cut design cycles and provide high complexity with maximum flexibility. These features, combined with lower package count and high reliability, truly make the PAL a circuit designer's best friend.

PAL<sup>®</sup> is a registered trademark of Monolithic Memories.

**The PAL—Teaching Old PROMs New Tricks**



MMI developed the modern PROM and introduced many of the architectures and techniques now regarded as industry standards. As the world's largest PROM manufacturer, MMI has the proven technology and high volume production capability required to manufacture and support the PAL.

The PAL is an extension of the fusible link technology pioneered by Monolithic Memories for use in bi-polar PROMs. The fusible link PROM first gave the digital systems designer the power to "write on silicon." In a few seconds he was able to transform a blank PROM from a general purpose device into one containing a custom algorithm, microprogram, or Boolean transfer function. This opened up new horizons for the use of PROMs in computer control stores, character generators, data storage tables and many other applications. The wide acceptance of this technology is clearly demonstrated by today's multi-million dollar PROM market.

The key to the PROM's success is that it allows the designer to quickly and easily customize the chip to fit his unique requirements. The PAL extends this programmable flexibility by utilizing proven fusible link technology to implement logic functions. Using PAL circuits the designer can quickly and effectively implement custom logic varying in complexity from random gates to complex arithmetic functions.

**ANDs and ORs**

The PAL implements the familiar sum of products logic by using a programmable AND array whose output terms feed a fixed OR

array. Since the sum of products form can express any Boolean transfer function, the PAL circuit uses are only limited by the number of terms available in the AND - OR arrays. PAL devices come in different sizes to allow for effective logic optimization.

Figure 1 shows the basic PAL structure for a two input, one output logic segment. The general logic equation for this segment is

$$\text{Output} = (I_1 + \bar{f}_1)(I_1 + \bar{f}_2)(I_2 + \bar{f}_3)(I_2 + \bar{f}_4) + (I_1 + \bar{f}_5)(I_1 + \bar{f}_6)(I_2 + \bar{f}_7)(I_2 + \bar{f}_8)$$

where the "f" terms represent the state of the fusible links in the PAL AND array. An unblown link represents a logic 1. Thus,

fuse blown,  $f = 0$

fuse intact,  $f = 1$

An unprogrammed PAL has all fuses intact.

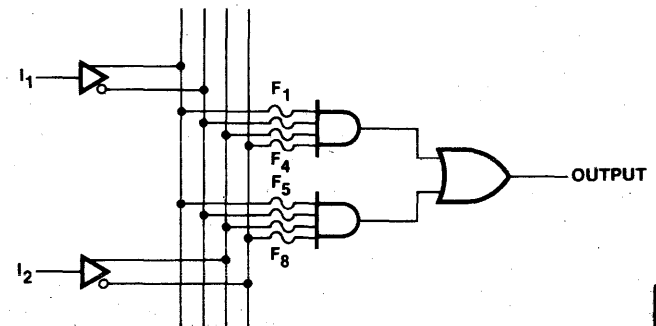


Figure 1

**PAL Notation**

Logic equations, while convenient for small functions, rapidly become cumbersome in large systems. To reduce possible confusion, complex logic networks are generally defined by logic diagrams and truth tables. Figure 2 shows the logic convention adopted to keep PAL logic easy to understand and use. In the figure, an "x" represents an intact fuse used to perform the logic AND function. (Note: the input terms on the common line with the x's are not connected together.) The logic symbology shown in Figure 2 has been informally adopted by integrated circuit manufacturers because it clearly establishes a one-to-one correspondence between the chip layout and the logic diagram. It also allows the logic diagram and truth table to be combined into a compact and easy to read form, thereby serving as a convenient shorthand for PAL circuits. The two input - one output example from Figure 1 redrawn using the new logic convention is shown in Figure 3.

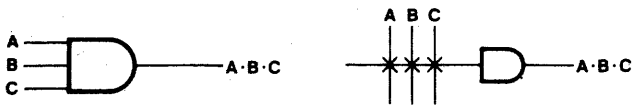


Figure 2

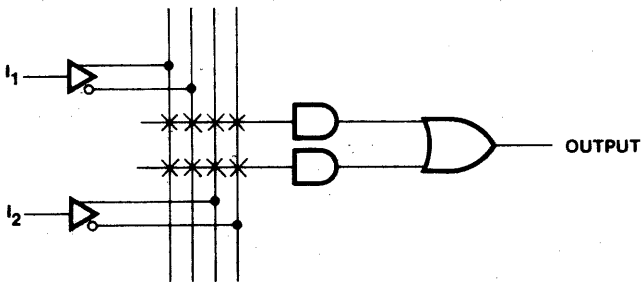


Figure 3

As a simple PAL example, consider the implementation of the transfer function:

$$\text{Output} = I_1 \bar{I}_2 + \bar{I}_1 I_2$$

The normal combinatorial logic diagram for this function is shown in figure 4, with the PAL logic equivalent shown in figure 5.

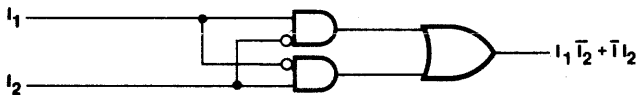


Figure 4

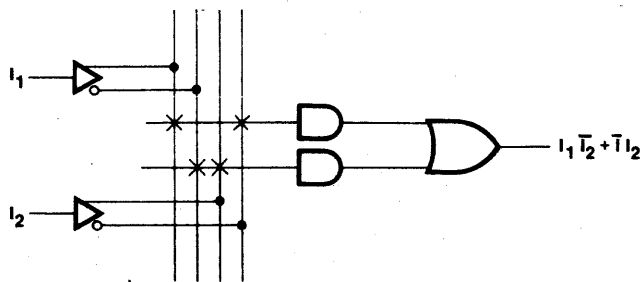


Figure 5

used to store computer programs and data. In these applications the fixed input is a computer memory address; the output is the contents of that memory location.

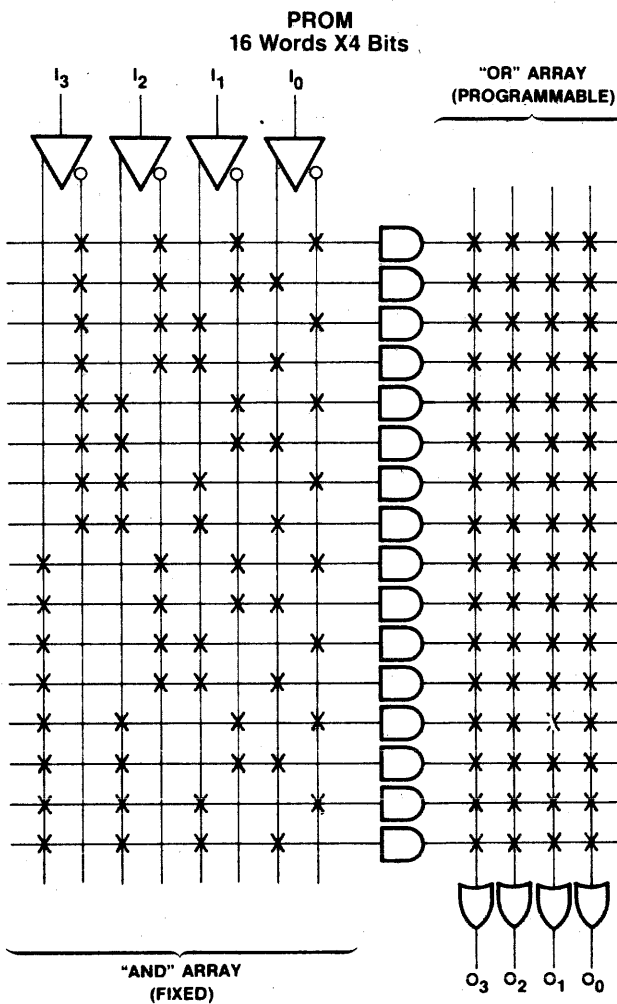


Figure 6

The basic logic structure of the PLA consists of a programmable AND array whose outputs feed a programmable OR array (Figure 7). Since the designer has complete control over all inputs and outputs, the PLA provides the ultimate flexibility for implementing logic functions. They are used in a wide variety of applications. However, this generality makes PLAs expensive, quite formidable to understand, and costly to program (they require special programmers).

The basic logic structure of the PAL, as mentioned earlier, consists of a programmable AND array whose outputs feed a fixed OR array (Figure 8). The PAL combines much of the flexibility of the PLA with the low cost and easy programmability of the PROM. Table 1 summarizes the characteristics of the PROM, PLA, and PAL logic families.

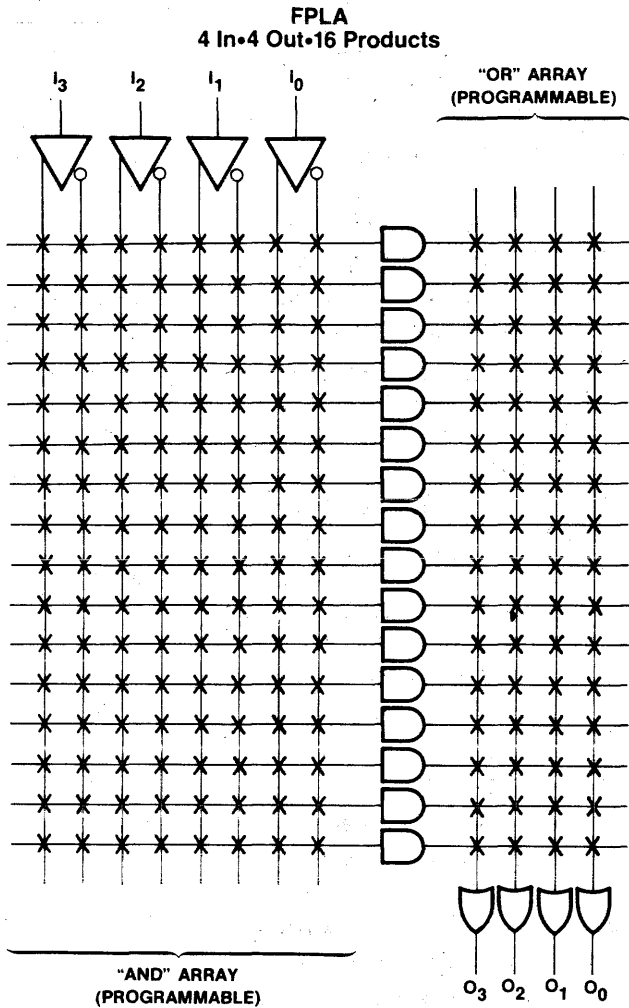


Figure 7

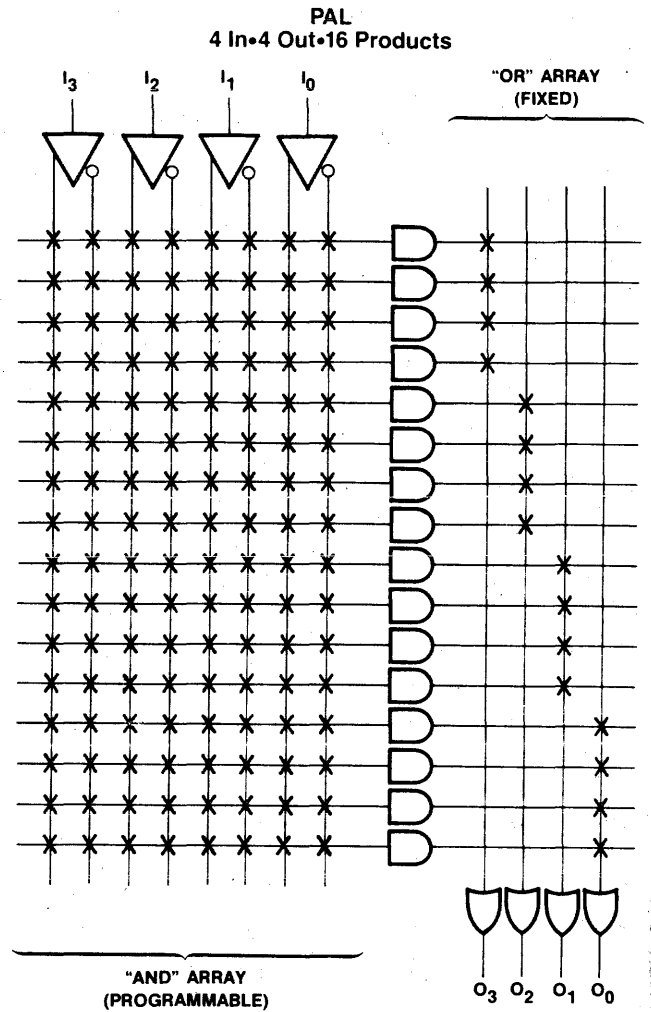


Figure 8

	AND	OR	OUTPUT OPTIONS
PROM	Fixed	Prog	TS, OC
FPLA	Prog	Prog	TS, OC, Fusible Polarity
FPGA	Prog	None	TS, OC, Fusible Polarity
FPLS	Prog	Prog	TS, Registered Feedback, I/O
PAL	Prog	Fixed	TS, Registered Feedback, I/O

Table 1

# PAL Introduction

## PAL Input/Output/Function/Performance Chart

PART NO.	INPUT	OUTPUT	PROGRAMMABLE I/O'S	FEEDBACK REGISTER	OUTPUT POLARITY	FUNCTIONS	PERFORMANCE			
							STD	A	-2	-4
10H8	10	8			AND-OR	AND-OR Gate Array	X		X	
12H6	12	6			AND-OR	AND-OR Gate Array	X		X	
14H4	14	4			AND-OR	AND-OR Gate Array	X		X	
16H2	16	2			AND-OR	AND-OR Gate Array	X		X	
16C1	16	2			BOTH <sup>1</sup>	AND-OR Gate Array	X		X	
20C1	20	2			BOTH <sup>1</sup>	AND-OR Gate Array	X			
10L8	10	8			AND-NOR	AND-OR Invert Gate Array	X		X	
12L6	12	6			AND-NOR	AND-OR Invert Gate Array	X		X	
14L4	14	4			AND-NOR	AND-OR Invert Gate Array	X		X	
16L2	16	2			AND-NOR	AND-OR Invert Gate Array	X		X	
12L10	12	10			AND-NOR	AND-OR Invert Gate Array	X			
14L8	14	8			AND-NOR	AND-OR Invert Gate Array	X			
16L6	16	6			AND-NOR	AND-OR Invert Gate Array	X			
18L4	18	4			AND-NOR	AND-OR Invert Gate Array	X			
20L2	20	2			AND-NOR	AND-OR Invert Gate Array	X			
16L8	10	2	6		AND-NOR	AND-OR Invert Gate Array	X	X	X	X
20L8	14	2	6		AND-NOR	AND-OR Invert Gate Array		X		
20L10	12	2	8		AND-NOR	AND-OR Invert Gate Array	X			
16R8	8	8		8	AND-NOR	AND-OR Invert Gate Array w/Reg's	X	X	X	X
16R6	8	6	2	6	AND-NOR	AND-OR Invert Array w/Reg's	X	X	X	X
16R4	8	4	4	4	AND-NOR	AND-OR Invert Array w/Reg's	X	X	X	X
20R8	12	8		8	AND-NOR	AND-OR Invert w/Reg's		X		
20R6	12	6	2	6	AND-NOR	AND-OR Invert w/Reg's		X		
20R4	12	4	4	4	AND-NOR	AND-OR Invert w/Reg's		X		
20X10	10	10		10	AND-NOR	AND-OR-XOR Invert w/Reg's	X			
20X8	10	8	2	8	AND-NOR	AND-OR-XOR Invert w/Reg's	X			
20X4	10	4	6	4	AND-NOR	AND-OR-XOR Invert w/Reg's	X			
16X4	8	4	4	4	AND-NOR	AND-OR-XOR Invert w/Reg's	X			
16A4	8	4	4	4	AND-NOR	AND-CARRY-OR-XOR Invert w/Reg's	X			

Table 2

<sup>1</sup> Simultaneous AND-OR and AND-NOR outputs

### PAL Circuits For Every Task

The members of the PAL family and their characteristics are summarized in Table 2. They are designed to cover the spectrum of logic functions at reduced cost and lower package count. This allows the designer to select the PAL that best fits his application. PAL units come in the following basic configurations:

### Gate Arrays

PAL gate arrays are available in sizes from 12x10 (12 input terms, 10 output terms) to 20x2, with both active high and active low output configurations available (figure 9). This wide variety of input/output formats allows the PAL to replace many different sized blocks of combinatorial logic with single packages.

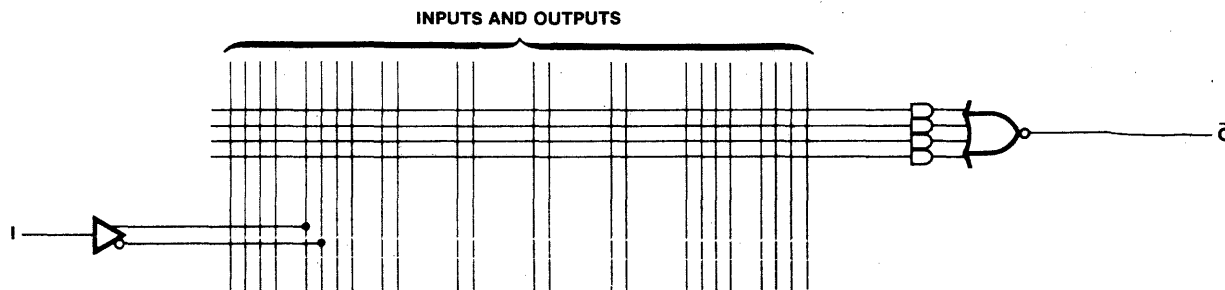


Figure 9

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## PAL Introduction

### Programmable I/O

A feature of the high-end members of the PAL family is programmable input/output. This allows the product terms to directly control the outputs of the PAL (Figure 10). One product term is used to enable the three-state buffer, which in turn gates the summation term to the output pin. The output is also fed

back into the PAL array as an input. Thus the PAL drives the I/O pin when the three-state gate is enabled; the I/O pin is an input to the PAL array when the three-state gate is disabled. This feature can be used to allocate available pins for I/O functions or to provide bi-directional output pins for operations such as shifting and rotating serial data.

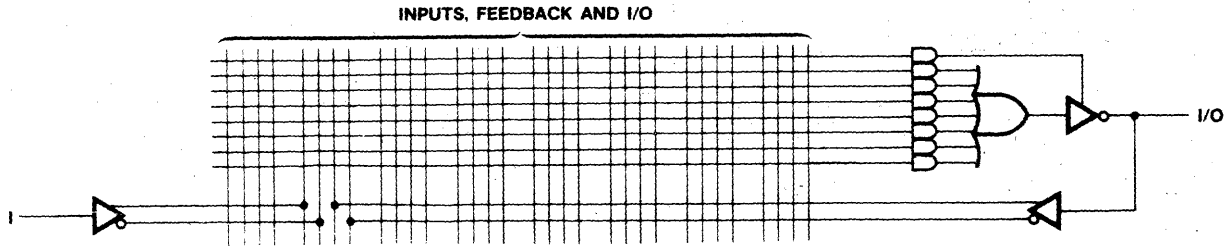


Figure 10

### Registered Outputs with Feedback

Another feature of the high end members of the PAL family is registered data outputs with registered feedback. Each product term is stored into a D-type output flip-flop on the rising edge of the system clock (Figure 11). The Q output of the flip-flop can then be gated to the output pin by enabling the active low three-state buffer.

In addition to being available for transmission, the Q output is fed back into the PAL array as an input term. This feedback allows the PAL to "remember" the previous state, and it can alter its function based upon that state. This allows the designer to configure the PAL as a state sequencer which can be programmed to execute such elementary functions as count up, count down, skip, shift, and branch. These functions can be executed by the registered PAL at rates of up to 25 MHz.

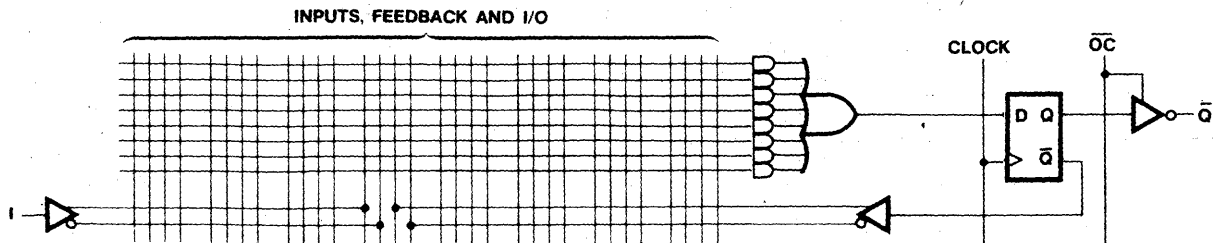


Figure 11

### XOR PALs

These PAL devices feature an exclusive OR function. The sum of products is segmented into two sums which are then exclusive ORed (XOR) at the input of the D-type flip-flop (Figure 12). All

of the features of the Registered PALs are included in the XOR PAL unit. The XOR function provides an easy implementation of the HOLD operation used in counters and other state sequencers.

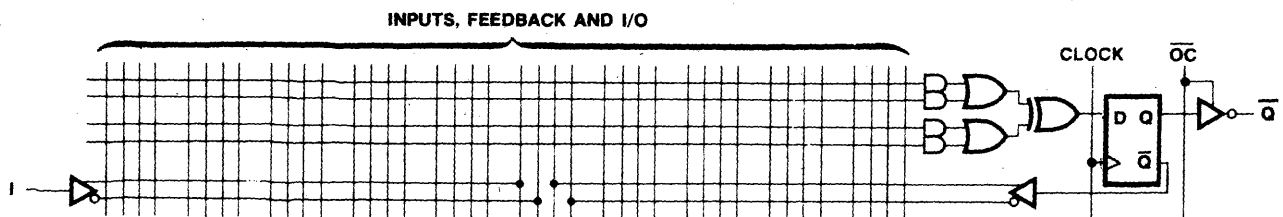


Figure 12



# PAL Introduction

## Arithmetic Gated Feedback

The arithmetic functions (add, subtract, greater than, and less than) are implemented by addition of gated feedback to the features of the XOR PAL device. The XOR at the input of the D-type flip-flop allows carries from previous operations to be XORed with two variable sums generated by the PAL array. The flip-flop Q output is fed back to be gated with input terms A

(Figure 13). This gated feedback provides any one of the possible Boolean combinations which are mapped in the Karnaugh map (Figure 15). Figure 14 shows how the PAL array can be programmed to perform these 16 operations. These features provide for versatile operations on two variables and facilitate the parallel generation of carries necessary for fast arithmetic operations.

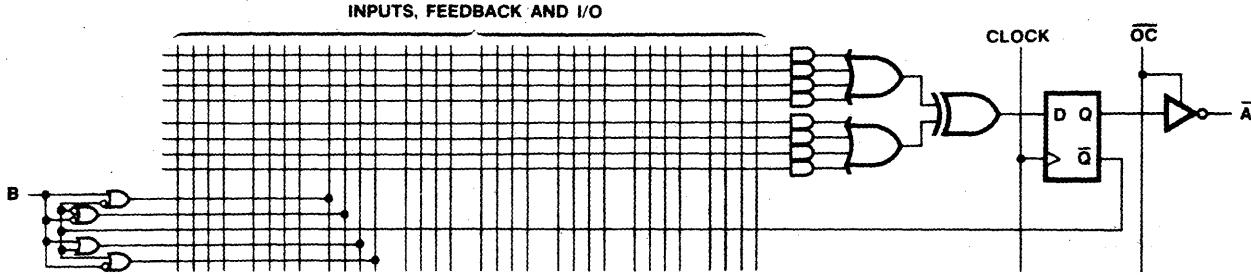
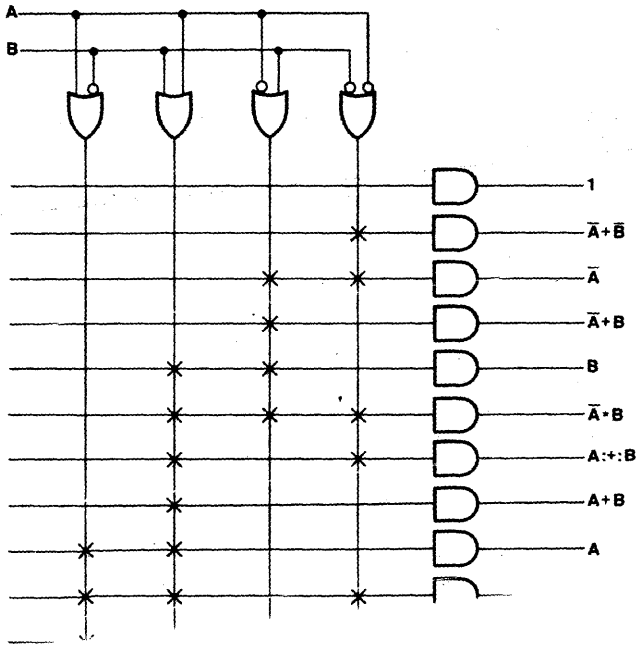


Figure 13

Monolithic Memories



$(\bar{A} + B) \cdot (\bar{A} + \bar{B})$   
 $(A + B) \cdot (A + B)$

	--	-x	xx	x-
--	1	$\bar{A} + \bar{B}$	$\bar{A}$	$\bar{A} + B$
-x	$A + B$	$A + B$	$\bar{A} \cdot B$	$B$
xx	$A$	$A \cdot \bar{B}$	0	$A \cdot B$

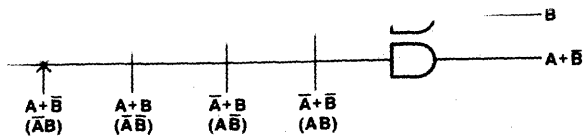


Figure 14

# PAL Introduction

## Programmable I/O

A feature of the high-end members of the PAL family is programmable input/output. This allows the product terms to directly control the outputs of the PAL (Figure 10). One product term is used to enable the three-state buffer, which in turn gates the summation term to the output pin. The output is also fed

back into the PAL array as an input. Thus the PAL drives the I/O pin when the three-state gate is enabled; the I/O pin is an input to the PAL array when the three-state gate is disabled. This feature can be used to allocate available pins for I/O functions or to provide bi-directional output pins for operations such as shifting and rotating serial data.

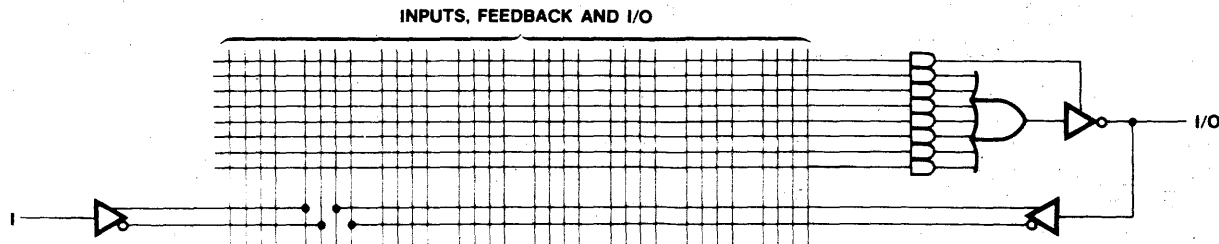


Figure 10

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Another feature of the high end members of the PAL family is registered data outputs with registered feedback. Each product term is stored into a D-type output flip-flop on the rising edge of the system clock (Figure 11). The Q output of the flip-flop can then be gated to the output pin by enabling the active low three-state buffer.

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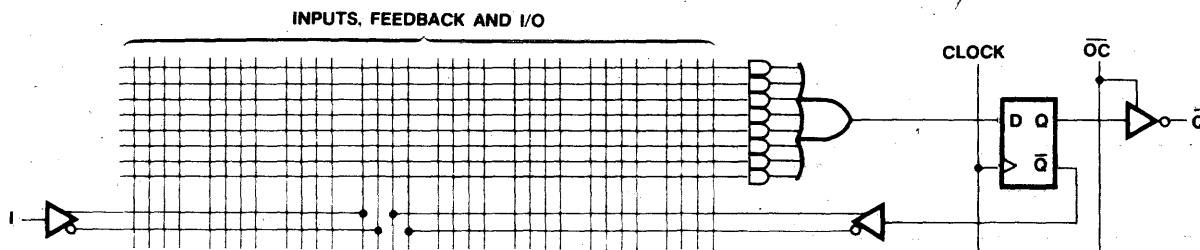


Figure 11

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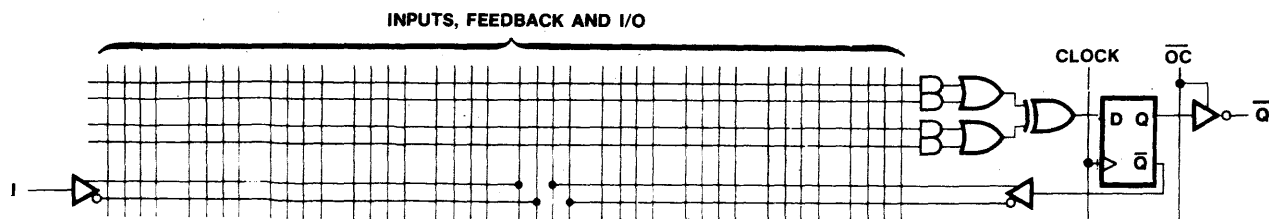


Figure 12

**Arithmetic Gated Feedback**

The arithmetic functions (add, subtract, greater than, and less than) are implemented by addition of gated feedback to the features of the XOR PAL device. The XOR at the input of the D-type flip-flop allows carries from previous operations to be XORed with two variable sums generated by the PAL array. The flip-flop Q output is fed back to be gated with input terms A

(Figure 13). This gated feedback provides any one of the 16 possible Boolean combinations which are mapped in the Karnaugh map (Figure 15). Figure 14 shows how the PAL array can be programmed to perform these 16 operations. These features provide for versatile operations on two variables and facilitate the parallel generation of carries necessary for fast arithmetic operations.

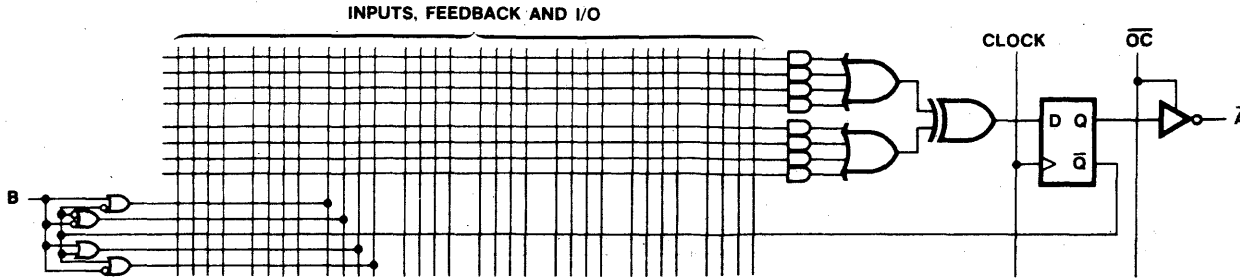


Figure 13

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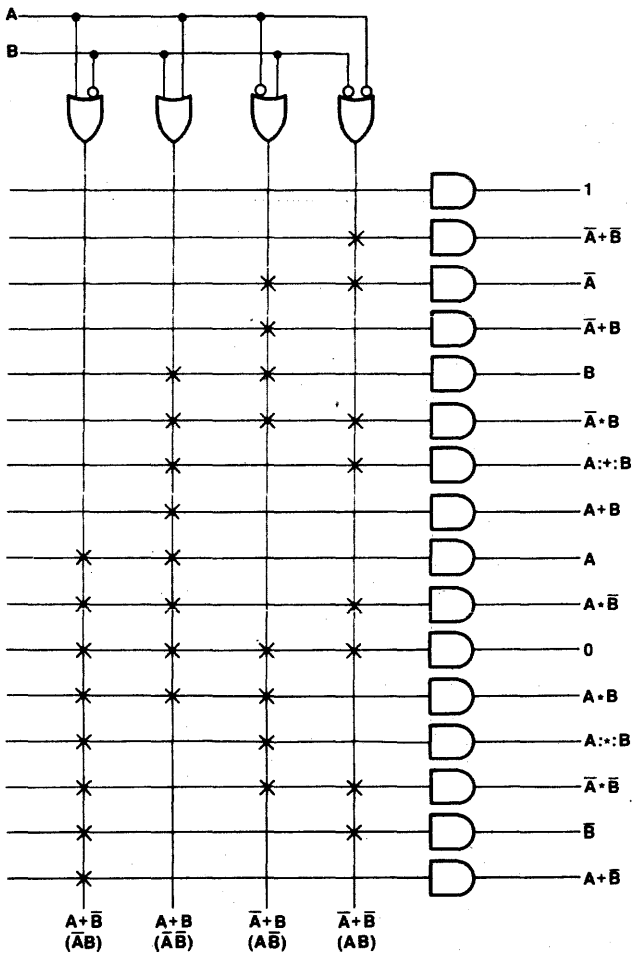


Figure 14

$(\bar{A}+B), (\bar{A}+B)$   
 $(A+B), (A+B)$

	--	-x	xx	x-
--	1	$\bar{A}+B$	$\bar{A}$	$\bar{A}+B$
-x	$A+B$	$A+B$	$\bar{A}+B$	B
xx	A	$A\cdot\bar{B}$	0	$A\cdot B$
x-	$A+B$	$\bar{B}$	$\bar{A}+B$	$A+B$

Figure 15

## PAL Introduction

It should now be clear that the PAL family can replace most Small-Scale Integrated Logic (SSI) logic in use today, thereby lowering product cost and giving the designer even greater flexibility in implementing logic functions.

### PAL Programming

PAL devices can be programmed in most standard PROM programmers with the addition of a PAL personality card. The PAL appears to the programmer as a PROM. During programming half of the PAL outputs are selected for programming while the other outputs and the inputs are used for addressing. The outputs are then switched to program the other locations. Verification uses the same procedure with the programming lines held in a low state.

### PALASM (PAL Assembler)

PALASM is the software used to define, simulate, build, and test PAL units. PALASM accepts the PAL Design Specification as an input file. It verifies the design against an optional function table and generates the fuse plot which is used to program the PAL devices. PALASM is available upon request for many computers and is documented in the PAL Design Concepts section.

### HAL (Hard Array Logic)

The HAL family is the mask programmed version of a PAL. The HAL is to a PAL just as a ROM is to a PROM. A standard wafer is fabricated to the 6th mask. Then a custom metal mask is used to fabricate Aluminum links for a HAL instead of the programmable Ti-W fuse array used in a PAL.

The HAL is a cost-effective solution for large quantities and is unique in that it is a gate array with a programmable prototype.

### HMSI (HAL Medium Scale Integration)

The HMSI family is derived from the PAL using HAL technology. These devices perform predetermined functions which are not available in the existing TTL family. Because they are produced in volume, the user receives the benefit of volume pricing. HMSI PAL designs are given in the Applications section with their industry standard 74LS part number in line 2 of the PAL Design Specification.

### PMSI (PAL Medium Scale Integration)

The PMSI family is derived in a similar fashion to HMSI except this product is produced entirely from a PAL circuit. A HAL circuit mask is not generated and an industry standard 74LS part number is not assigned unless sales warrant it.

### PAL Technology

PAL circuits are manufactured using the proven TTL Schottky bipolar Ti-W fuse process to make fusible-link PROMs. An NPN emitter follower array forms the programmable AND array. PNP inputs provide high-impedance inputs (0.25 mA max) to the array. All outputs are standard TTL drivers with internal active pull-up transistors. Typical PAL propagation delay time is 25 ns, and all PALs are packaged in space saving 20-pin and 24-pin SKINNYDIP® packages.

### PAL Data Security

The circuitry used for programming and logic verification can be used at any time to determine the logic pattern stored in the PAL array. For security, the PAL has a "last fuse" which can be blown to disable the verification logic. This provides a significant deterrent to potential copiers, and it can be used to effectively protect proprietary designs.

# PAL<sup>®</sup>-Programmable Array Logic

## HAL<sup>®</sup>-Hard Array Logic

### Features/Benefits

- Reduces SSI/MSI chip count greater than 5 to 1
- Saves space with SKINNYDIP<sup>®</sup> packages
- Reduces IC inventories substantially
- Expedites and simplifies prototyping and board layout
- PALASM<sup>™</sup> silicon compiler provides auto routing and test vectors
- Security fuse reduces possibility of copying by competitors

### Description

The PAL family utilizes an advanced Schottky TTL process and the Bipolar PROM fusible link technology to provide user programmable logic for replacing conventional SSI/MSI gates and flip-flops at reduced chip count.

The HAL family utilizes standard Low-Power Schottky TTL process and automated mask pattern generation directly from logic equations to provide a semi-custom gate array for replacing conventional SSI/MSI gates and flip-flops at reduced chip count.

There are four different speed/power families offered. Choose from either the standard, high speed, half power, or quarter power family to maximize design performance.

The PAL/HAL lets the systems engineer "design his own chip" by blowing fusible links to configure AND and OR gates to perform his desired logic function. Complex interconnections which previously required time-consuming layout are thus "lifted" from PC board etch and placed on silicon where they can be easily modified during prototype check-out or production.

The PAL transfer function is the familiar sum of products. Like the PROM, the PAL has a single array of fusible links. Unlike the PROM, the PAL is a programmable AND array driving a fixed OR array (the PROM is a fixed AND array driving a programmable OR array).

The HAL transfer function is the familiar sum of products. Like the ROM, the HAL has a single array of selectable gates. Unlike the ROM, the HAL is a selectable AND array driving a fixed OR array (the ROM is a fixed AND array driving a selectable OR array).

In addition the PAL/HAL provides these options:

- Variable input/output pin ratio
- Programmable three-state outputs
- Registers with feedback
- Arithmetic capability
- Exclusive-OR gates

PAL<sup>®</sup>, (Programmable Array Logic), PALASM<sup>®</sup>, HAL<sup>®</sup>, and SKINNYDIP<sup>®</sup> are registered trademarks and PMSI, and HMSI are trademarks of Monolithic Memories Inc.

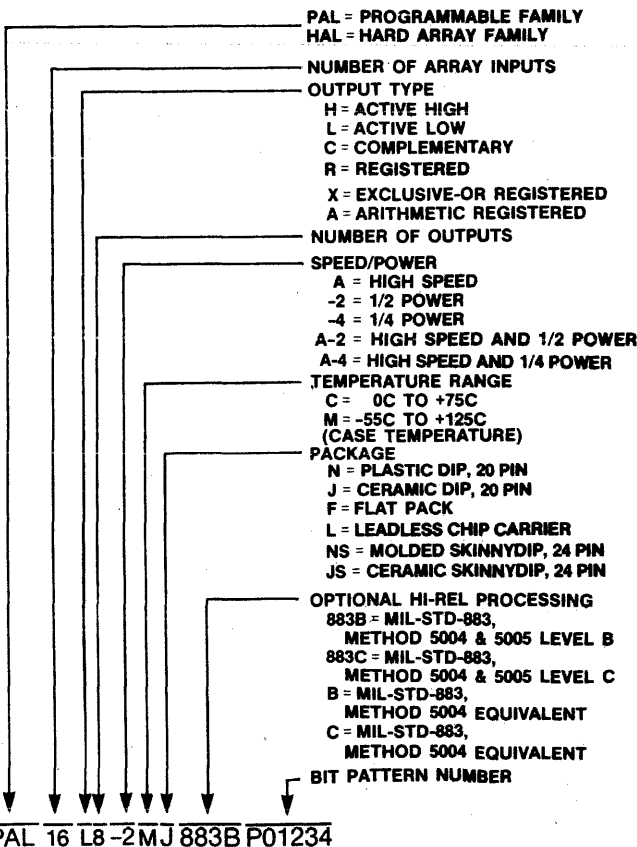
Unused inputs are tied directly to V<sub>CC</sub> or GND. Product terms with all fuses blown assume the logical high state, and product terms connected to both true and complement of any single input assume the logical low state. Registers consist of D type flip-flops which are loaded on the low-to-high transition of the clock. PAL/HAL Logic Diagrams are shown with all fuses blown, enabling the designer to use the diagrams as coding sheets.

The entire PAL family is programmed using inexpensive conventional PROM programmers with appropriate personality and socket adapter cards. Once the PAL is programmed and verified, two additional fuses may be blown to defeat verification. This feature gives the user a proprietary circuit which is very difficult to copy.

To design a HAL, the user first programs and debugs a PAL using PALASM and the "PAL DESIGN SPECIFICATION" standard format. This specification is submitted to Monolithic Memories where it is computer processed and assigned a bit pattern number, e.g., P01234.

Monolithic Memories will provide a PAL sample for customer qualification. The user then submits a purchase order for a HAL of the specified bit pattern number, e.g., HAL18L4 P01234. See Ordering Information below.

### Ordering Information



TWX: 910-338-2376  
 TWX: 910-338-2374

2175 Mission College Boulevard, Santa Clara, CA 95050 Tel: (408) 970-9700

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## 20/24-Pin PAL/HAL

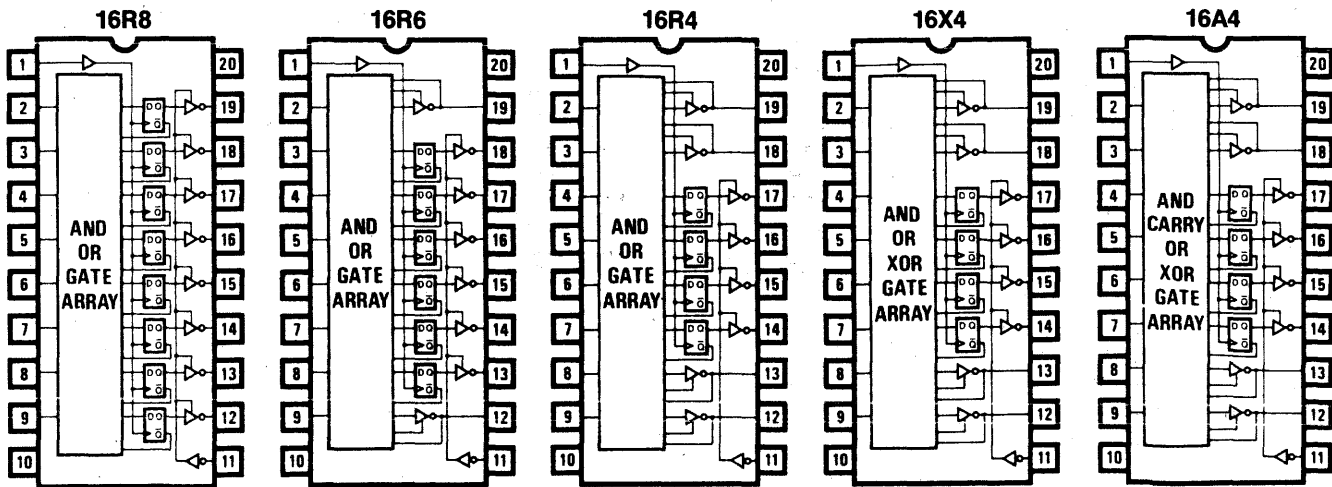
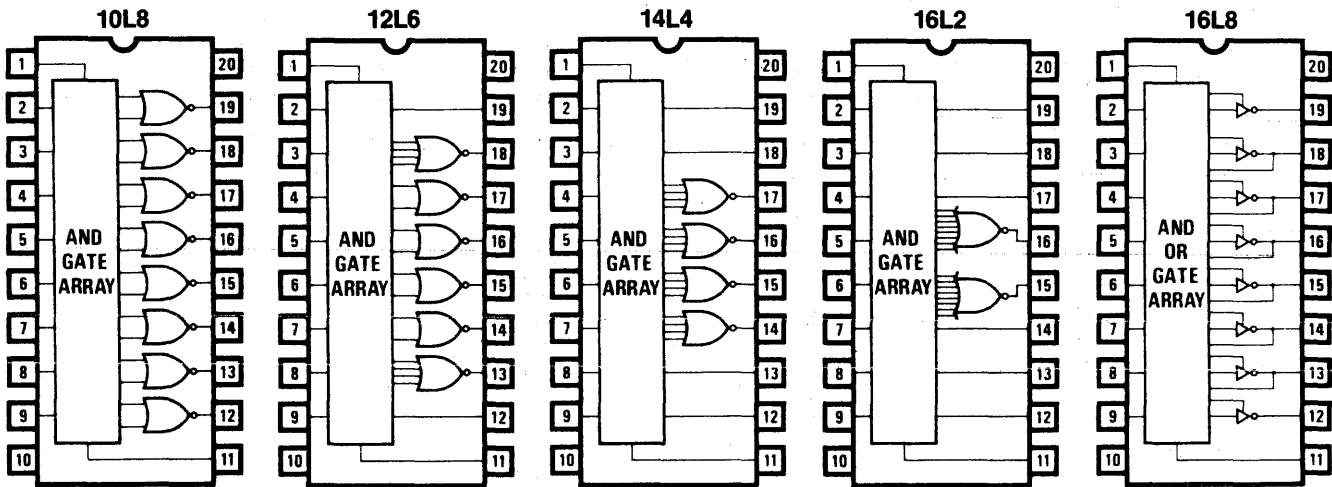
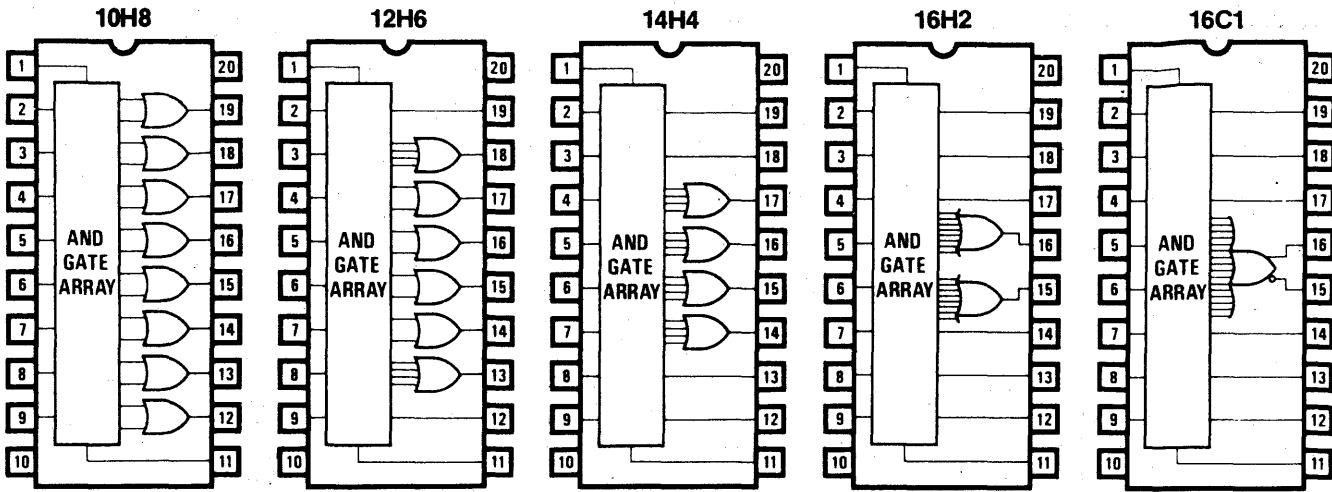
GENERIC LOGIC	PINS	PACKAGE	DESCRIPTION	PART NUMBER			
				STANDARD	HIGH SPEED	1/2 POWER	1/4 POWER
10H8	20	N, J, F, L	Octal 10 Input And-Or Gate Array	PAL10H8 HAL10H8		PAL10H8-2 HAL10H8-2	
12H6	20	N, J, F, L	Hex 12 Input And-Or Gate Array	PAL12H6 HAL12H6		PAL12H6-2 HAL12H6-2	
14H4	20	N, J, F, L	Quad 14 Input And-Or Gate Array	PAL14H4 HAL14H4		PAL14H4-2 HAL14H4-2	
16H2	20	N, J, F, L	Dual 16 Input And-Or Gate Array	PAL16H2 HAL16H2		PAL16H2-2 HAL16H2-2	
16C1	20	N, J, F, L	16 Input And-Or/And-Or-Invert Gate Array	PAL16C1 HAL16C1		PAL16C1-2 PAL16C1-2	
10L8	20	N, J, F, L	Octal 10 Input And-Or-Invert Gate Array	PAL10L8 HAL10L8		PAL10L8-2 HAL10L8-2	
12L6	20	N, J, F, L	Hex 12 Input And-Or-Invert Gate Array	PAL12L6 HAL12L6		PAL12L6-2 HAL12L6-2	
14L4	20	N, J, F, L	Quad 14 Input And-Or-Invert Gate Array	PAL14L4 HAL14L4		PAL14L4-2 HAL14L4-2	
16L2	20	N, J, F, L	Dual 16 Input And-Or-Invert Gate Array	PAL16L2 HAL16L2		PAL16L2-2 HAL16L2-2	
16L8	20	N, J, F, L	Octal 16 Input And-Or-Invert Gate Array	PAL16L8 * HAL16L8	PAL16L8A HAL16L8A	PAL16L8A-2 HAL16L8A-2	PAL16L8A-4 HAL16L8A-4
16R8	20	N, J, F, L	Octal 16 Input Registered And-Or Gate Array	PAL16R8 * HAL16R8	PAL16R8A HAL16R8A	PAL16R8A-2 HAL16R8A-2	PAL16R8A-4 HAL16R8A-4
16R6	20	N, J, F, L	Hex 16 Input Registered And-Or Gate Array	PAL16R6 * HAL16R6	PAL16R6A HAL16R6A	PAL16R6A-2 HAL16R6A-2	PAL16R6A-4 HAL16R6A-4
16R4	20	N, J, F, L	Quad 16 Input Registered And-Or Gate Array	PAL16R4 * HAL16R4	PAL16R4A HAL16R4A	PAL16R4A-2 HAL16R4A-2	PAL16R4A-4 HAL16R4A-4
16X4	20	N, J, F, L	Quad 16 Input Registered And-Or-Xor Gate Array	PAL16X4 HAL16X4			
16A4	20	N, J, F, L	Quad 16 Input Registered And-Carry-Or-Xor Gate Array	PAL16A4 HAL16A4			
12L10	24 (28)	NS,JS,F (L)	Deca 12 Input And-Or-Invert Gate Array	PAL12L10 HAL12L10			
14L8	24 (28)	NS,JS,F (L)	Octal 14 Input And-Or-Invert Gate Array	PAL14L8 HAL14L8			
16L6	24 (28)	NS,JS,F (L)	Hex 16 Input And-Or-Invert Gate Array	PAL16L6 HAL16L6			
18L4	24 (28)	NS,JS,F (L)	Quad 18 Input And-Or-Invert Gate Array	PAL18L4 HAL18L4			
20L2	24 (28)	NS,JS,F (L)	Dual 20 Input And-Or-Invert Gate Array	PAL20L2 HAL20L2			
20C1	24 (28)	NS,JS,F (L)	20 Input And-Or/And-Or-Invert Gate Array	PAL20C1 HAL20C1			
20L10	24 (28)	NS,JS,F (L)	Deca 20 Input And-Or-Invert Gate Array	PAL20L10 HAL20L10			
20X10	24 (28)	NS,JS,F (L)	Deca 20 Input Registered And-Or-Xor Gate Array	PAL20X10 HAL20X10			
20X8	24 (28)	NS,JS,F (L)	Octal 20 Input Registered And-Or-Xor Gate Array	PAL20X8 HAL20X8			
20X4	24 (28)	NS,JS,F (L)	Quad 20 Input Registered And-Or-Xor Gate Array	PAL20X4 HAL20X4			
20L8	24 (28)	NS,JS,F (L)	Octal 20 Input And-Or-Invert Gate Array		PAL20L8A HAL20L8A		
20R8	24 (28)	NS,JS,F (L)	Octal 20 Input Registered And-Or Gate Array		PAL20R8A HAL20R8A		
20R6	24 (28)	NS,JS,F (L)	Hex 20 Input Registered And-Or Gate Array		PAL20R6A HAL20R6A		
20R4	24 (28)	NS,JS,F (L)	Quad 20 Input Registered And-Or Gate Array		PAL20R4A HAL20R4A		

\* HAL available in "W" package. ( ) = Military Product Standard.

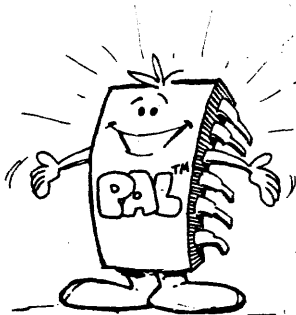
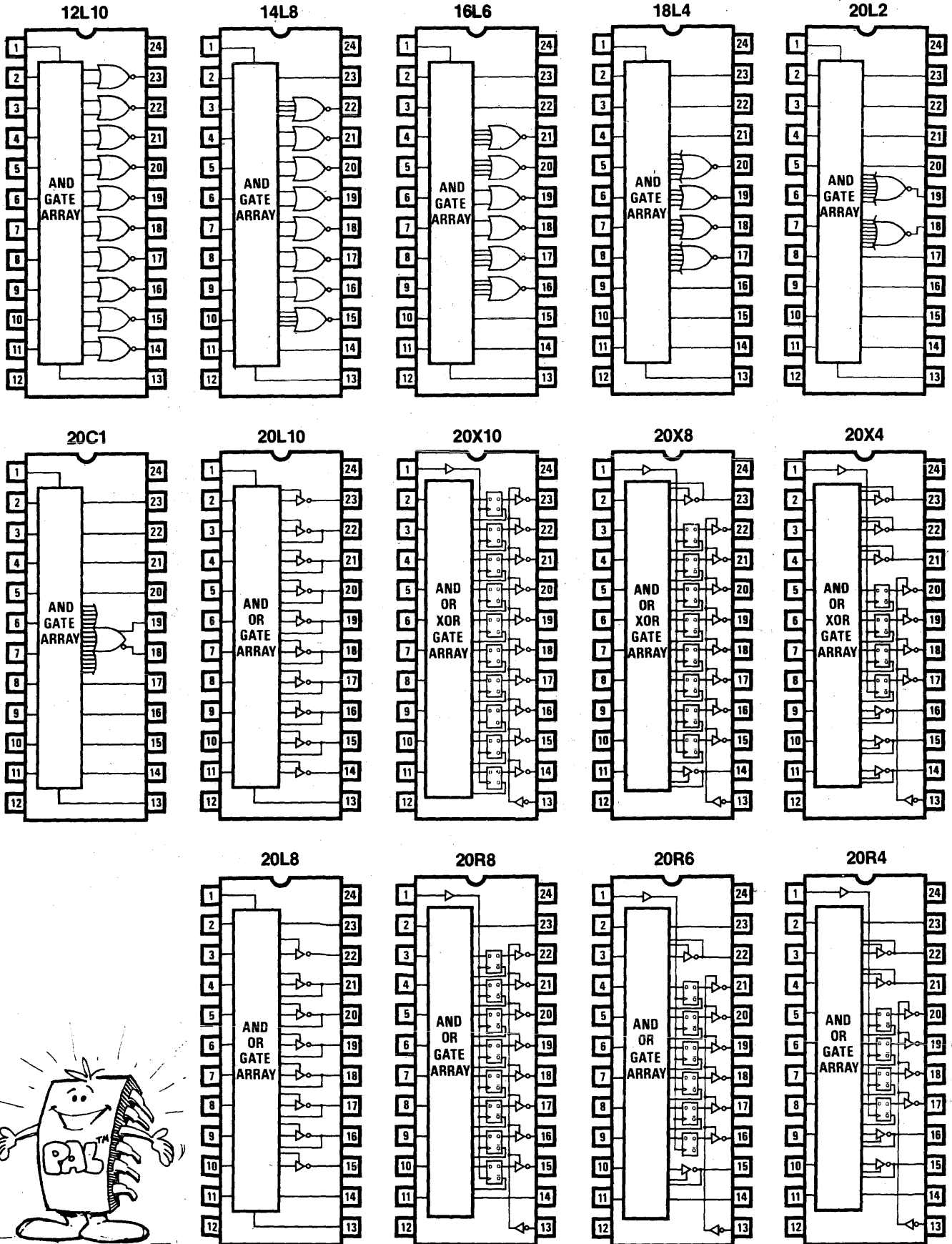
# 20-Pin PAL/HAL

Monolithic Memories

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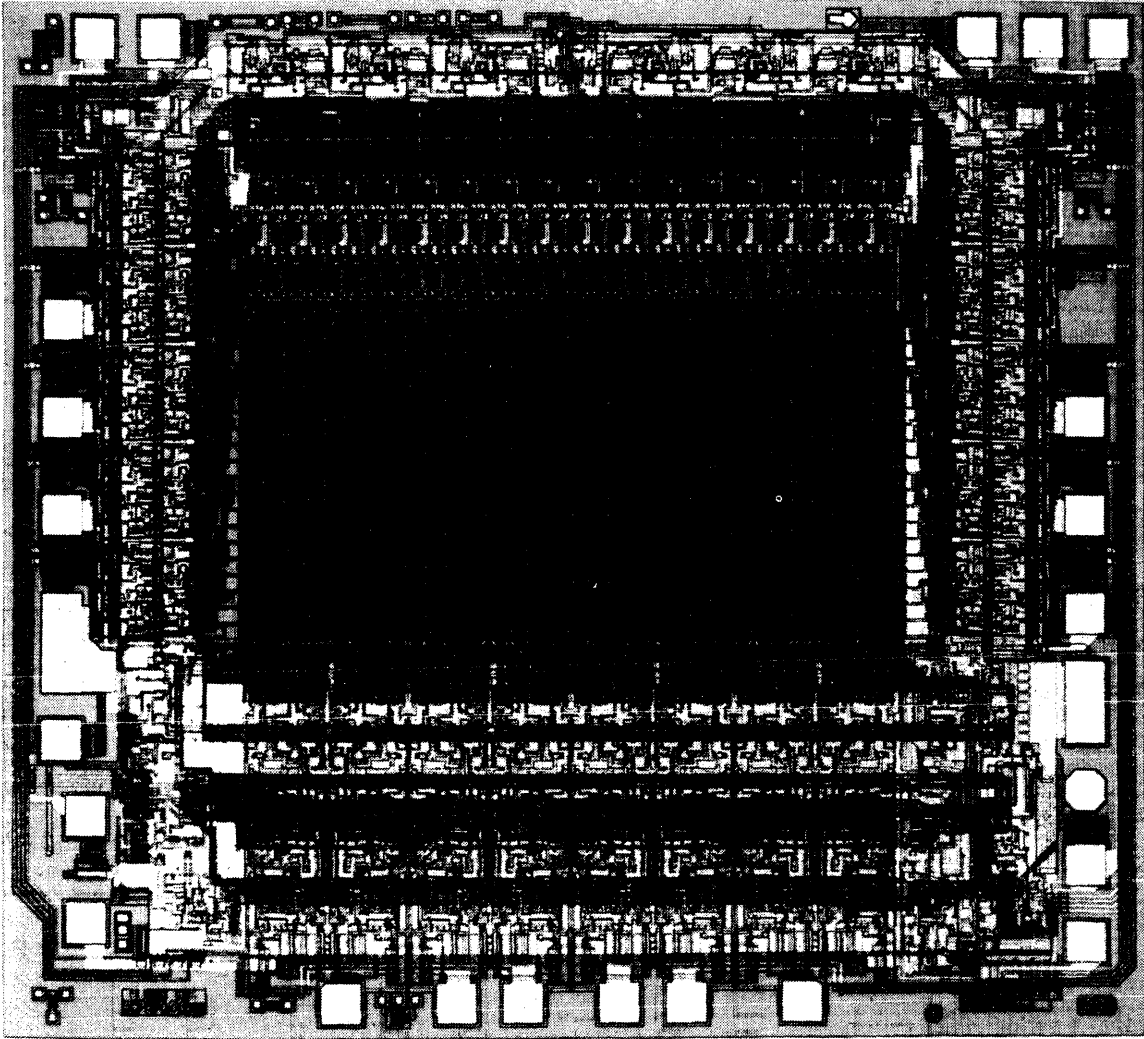
# 24-Pin PAL/HAL



Monolithic Memories

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Monolithic Memories

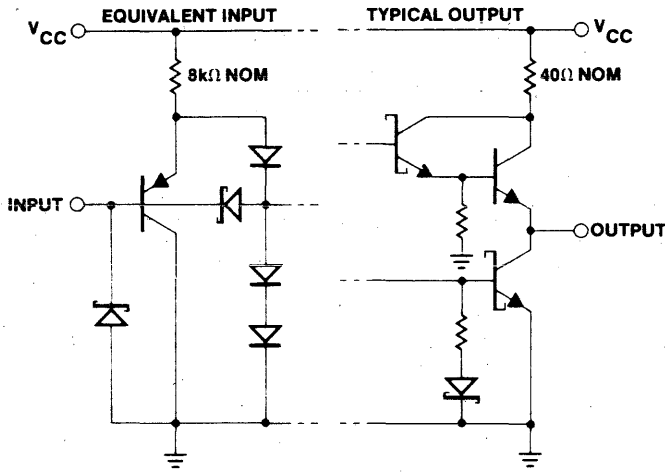
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**Absolute Maximum Ratings**

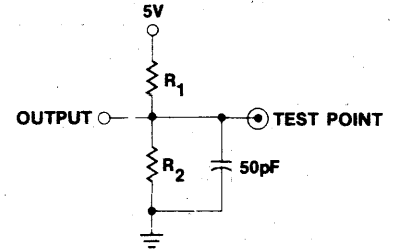
	Operating	Programming
Supply Voltage, $V_{CC}$ .....	-0.5 to 7.0V	-0.5 to 12.0V
Input Voltage .....	-1.5 to 5.5V	-1.0 to 22V ⊕
Off-state output Voltage .....	5.5V	12.0V
Storage temperature .....		-65° to +150°C

⊕ Pins 1 and 11 may be raised to 20V

**Schematic of Inputs and Outputs**



**Test Load**



Other loads may be used.

**Typical notes for all the following specifications**

Notes: Apply to electrical and switching characteristics

- † I/O pin leakage is the worst case of  $I_{OZX}$  or  $I_{IX}$  e.g.,  $I_{IL}$  and  $I_{OZH}$ .
- \* These are absolute voltages with respect to the ground pin on the device and includes all overshoots due to system and/or tester noise. Do not attempt to test these values without suitable equipment.
- \*\* Only one output shorted at a time.

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
T <sub>A</sub>	Operating free-air temperature	-55			0		75	°C
T <sub>C</sub>	Operating case temperature			125				°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS				UNIT
			MIN	TYP	MAX	
V <sub>IL</sub> *	Low-level input voltage				0.8	V
V <sub>IH</sub> *	High-level input voltage				2	V
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN I <sub>I</sub> = -18mA			-0.8 -1.5	V
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX V <sub>I</sub> = 0.4V			-0.02 -0.25	mA
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX V <sub>I</sub> = 2.4V			25	μA
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX V <sub>I</sub> = 5.5V			1	mA
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 8mA	0.3 0.5	V
			COM	I <sub>OL</sub> = 8mA		
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -2mA	2.4 2.8	V
			COM	I <sub>OH</sub> = -3.2mA		
I <sub>OS</sub>	Output short-circuit current**	V <sub>CC</sub> = 5V V <sub>O</sub> = 0V			-30 -70 -130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX			55 90	mA

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feed-back to output	Except 16C1	R1 = 560Ω R2 = 1.1kΩ	25	45	25	35	ns	
		16C1		25	45	25	40		

Monolithic Memories

CUSTOM/SEMICUSTOM

**Standard PAL/HAL Series 24**  
**12L10, 14L8, 16L6, 18L4, 20L2, 20C1**

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
T <sub>A</sub>	Operating free-air temperature	-55			75			°C
T <sub>C</sub>	Operating case temperature				125			°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
V <sub>IL</sub> *	Low-level input voltage					0.8	V
V <sub>IH</sub> *	High-level input voltage			2			V
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA	-0.8	-1.5		V
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V	-0.02	-0.25		mA
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V		25		μA
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V		1		mA
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 8mA	0.3	0.5	V
			COM	I <sub>OL</sub> = 8mA			
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -2mA	2.4	2.8	V
			COM	I <sub>OH</sub> = -3.2mA			
I <sub>OS</sub>	Output short-circuit current **	V <sub>CC</sub> = 5V	V <sub>O</sub> = 0V	-30	-70	-130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX			60	100	mA

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	R1 = 560Ω R2 = 1.1kΩ		25	45		25	40	ns

**Standard PAL/HAL Series 20**  
**16L8, 16R8, 16R6, 16R4, 16X4, 16A4**

**Operating Conditions**

SYMBOL	PARAMETER		MILITARY			COMMERCIAL			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage		4.5	5	5.5	4.75	5	5.25	V
t <sub>w</sub>	Width of clock	Low	25	10		25	10		ns
		High	25	10		25	10		
t <sub>su</sub>	Set up time from input or feedback to clock	16R8 16R6 16R4	45	25		35	25		ns
		16X4 16A4	55	30		45	30		
t <sub>h</sub>	Hold time		0	-15		0	-15		ns
T <sub>A</sub>	Operating free-air temperature		-55			0 75			°C
T <sub>C</sub>	Operating case temperature					125			°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN TYP MAX			UNIT
				MIN	TYP	MAX	
V <sub>IL</sub> *	Low-level input voltage			0.8			V
V <sub>IH</sub> *	High-level input voltage			2			V
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA	-0.8	-1.5		V
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V	-0.02	-0.25		mA
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V	25			μA
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V	1			mA
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL I <sub>OL</sub> = 12mA	0.3	0.5		V
			COM I <sub>OL</sub> = 24mA				
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL I <sub>OH</sub> = -2mA	2.4	2.8		V
			COM I <sub>OH</sub> = -3.2mA				
I <sub>OZL</sub>	Off-state output current †	V <sub>CC</sub> = MAX	V <sub>O</sub> = 0.4V	-100			μA
I <sub>OZH</sub>			V <sub>O</sub> = 2.4V	100			μA
I <sub>OS</sub>	Output short-circuit current **	V <sub>CC</sub> = 5V	V <sub>O</sub> = 0V	-30	-70	-130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX	16R4 16R6 16R8 16L8	120	180		mA
			16X4	160	225		
			16A4	170	240		

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER		TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	16R6 16R4 16L8	R <sub>1</sub> = 200Ω R <sub>2</sub> = 390Ω	25	45		25	35		ns
		16X4 16A4		30	45		30	40		ns
t <sub>CLK</sub>	Clock to output or feedback	15		25		15	25		ns	
t <sub>PZX</sub>	Pin 11 to output enable except 16L8	15		25		15	25		ns	
t <sub>PXZ</sub>	Pin 11 to output disable except 16L8	15		25		15	25		ns	
t <sub>PZX</sub>	Input to output enable	16R6 16R4 16L8		25	45		25	35		ns
		16X4 16A4		30	45		30	40		ns
t <sub>PXZ</sub>	input to output disable	16R6 16R4 16L8		25	45		25	35		ns
		16X4 16A4		30	45		30	40		ns
f <sub>MAX</sub>	Maximum frequency	16R8 16R6 16R4		14	25		16	25		MHz
		16X4 16A4	12	22		14	22			

**Standard PAL/HAL Series 24**  
**20X10, 20X8, 20X4, 20L10**

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
t <sub>w</sub>	Width of clock	Low	40	20	35	20		ns
		High	30	10	25	10		
t <sub>su</sub>	Set up time from input or feedback to clock	60	38		50	38		ns
t <sub>h</sub>	Hold time	0	-15		0	-15		ns
T <sub>A</sub>	Operating free-air temperature	-55			0		75	°C
T <sub>C</sub>	Operating case temperature			125				°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN TYP MAX			UNIT	
				MIN	TYP	MAX		
V <sub>IL</sub> *	Low-level input voltage					0.8	V	
V <sub>IH</sub> *	High-level input voltage					2	V	
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA			-0.8 -1.5	V	
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V			-0.02 -0.25	mA	
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V			25	μA	
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V			1	mA	
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 12mA	0.3	0.5	V	
			COM	I <sub>OL</sub> = 24mA				
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -2mA	2.4	2.8	V	
			COM	I <sub>OH</sub> = -3.2mA				
I <sub>OZL</sub>	Off-state output current †	V <sub>CC</sub> = MAX		V <sub>O</sub> = 0.4V			-100	μA
I <sub>OZH</sub>				V <sub>O</sub> = 2.4V			100	μA
I <sub>OS</sub>	Output short-circuit current**	V <sub>CC</sub> = 5V		V <sub>O</sub> = 0V			-30 -70 -130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX	20X10	20X8	20X4	120	180	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX	20L10			90	165	mA

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER		TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feed-back to output		R <sub>1</sub> = 200Ω R <sub>2</sub> = 390Ω		35	60	35	50	ns	
t <sub>CLK</sub>	Clock to output or feedback				20	35	20	30	ns	
t <sub>PXZ/ZX</sub>	Pin 13 to output disable/enable except 20L10				20	45	20	35	ns	
t <sub>PZX</sub>	Input to output enable except 20X10				35	55	35	45	ns	
t <sub>PXZ</sub>	Input to output disable except 20X10				35	55	35	45	ns	
f <sub>MAX</sub>	Maximum frequency				10.5	16	12.5	16	MHz	

**Fast PAL/HAL Series 20A**  
**16L8A, 16R8A, 16R6A, 16R4A**

**Operating Conditions**

SYMBOL	PARAMETER		MILITARY			COMMERCIAL			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage		4.5	5	5.5	4.75	5	5.25	V
t <sub>w</sub>	Width of clock	Low	20	10		15	10		ns
		High	20	10		15	10		
t <sub>su</sub>	Set up time from input or feedback to clock	16R8A 16R6A 16R4A	30	15		25 †	15		ns
t <sub>h</sub>	Hold time		0	-10		0	-10		ns
T <sub>A</sub>	Operating free-air temperature		-55			0 75			°C
T <sub>C</sub>	Operating case temperature					125			°C

†Can select 20ns upon customer request.

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN TYP MAX			UNIT
				MIN	TYP	MAX	
V <sub>IL</sub> *	Low-level input voltage					0.8	V
V <sub>IH</sub> *	High-level input voltage					2	V
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA			-0.8 -1.5	V
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V			-0.02 -0.25	mA
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V			25	μA
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V			1	mA
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 12mA		0.3 0.5	V
			COM	I <sub>OL</sub> = 24mA			
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -2mA	2.4	2.8	V
			COM	I <sub>OH</sub> = -3.2mA			
I <sub>OZL</sub>	Off-state output current †	V <sub>CC</sub> = MAX		V <sub>O</sub> = 0.4V		+100	μA
I <sub>OZH</sub>				V <sub>O</sub> = 2.4V		100	μA
I <sub>OS</sub>	Output short-circuit current **	V <sub>CC</sub> = 5V		V <sub>O</sub> = 0V	-30	-70 -130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX				120 180	mA

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER		TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	16R6A 16R4A 16L8A	R <sub>1</sub> = 200Ω R <sub>2</sub> = 390Ω	15	30		15	25	ns	
t <sub>CLK</sub>	Clock to output or feedback			10	20		10	15	ns	
t <sub>pZX</sub>	Pin 11 to output enable except 16L8A			10	25		10	20	ns	
t <sub>pXZ</sub>	Pin 11 to output disable except 16L8A			11	25		11	20	ns	
t <sub>pZX</sub>	Input to output enable	16R6A 16R4A 16L8A		10	30		10	25	ns	
t <sub>pXZ</sub>	Input to output disable	16R6A 16R4A 16L8A		13	30		13	25	ns	
f <sub>MAX</sub>	Maximum frequency	16R8A 16R6A 16R4A		20	40		28.5	40	MHz	

TWX: 910-338-2376

2175 Mission College Boulevard, Santa Clara, CA 95050 Tel: (408) 970-9700

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**Fast Series 24A**  
**20L8A, 20R8A, 20R6A, 20R4A**

**Operating Conditions**

SYMBOL	PARAMETER		MILITARY			COMMERCIAL			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage		4.5	5	5.5	4.75	5	5.25	V
t <sub>w</sub>	Width of clock	Low	20	7		15	7		ns
		High	20	7		15	7		
t <sub>su</sub>	Set up time from input or feedback to clock	20R8A 20R6A 20R4A	30	15		25	15		ns
t <sub>h</sub>	Hold time		0	-10		0	-10		ns
T <sub>A</sub>	Operating free-air temperature		-55			0 75			°C
T <sub>C</sub>	Operating case temperature					125			°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
V <sub>IL</sub> *	Low-level input voltage					0.8	V	
V <sub>IH</sub> *	High-level input voltage			2			V	
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA	-0.8	-1.5		V	
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V	-0.02	-0.25		mA	
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V		25		μA	
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V		1		mA	
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 12mA	0.3	0.5	V	
			COM	I <sub>OL</sub> = 24mA				
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -2mA	2.4	2.8	V	
			COM	I <sub>OH</sub> = -3.2mA				
I <sub>OZL</sub>	Off-state output current †	V <sub>CC</sub> = MAX		V <sub>O</sub> = 0.4V		-100	μA	
I <sub>OZH</sub>				V <sub>O</sub> = 2.4V		100	μA	
I <sub>OS</sub>	Output short-circuit current **	V <sub>CC</sub> = 5V		V <sub>O</sub> = 0V	-30	-90	-130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX			160	210	mA	

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER		TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	20R6A 20R4A 20L8A	R <sub>1</sub> = 200Ω R <sub>2</sub> = 390Ω	15	30		15	25	ns	
t <sub>CLK</sub>	Clock to output or feedback			10	20		10	15	ns	
t <sub>PZX</sub>	Pin 13 to output enable except 20L8A			10	25		10	20	ns	
t <sub>PXZ</sub>	Pin 13 to output disable except 20L8A			11	25		11	20	ns	
t <sub>PZX</sub>	Input to output enable	20R6A 20R4A 20L8A		10	30		10	25	ns	
t <sub>PXZ</sub>	Input to output disable	20R6A 20R4A 20L8A		13	30		13	25	ns	
f <sub>MAX</sub>	Maximum frequency	20R8A 20R6A 20R4A		20	40		28.5	40	MHz	



**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
T <sub>A</sub>	Operating free-air temperature	-55		125	0		75	°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN TYP MAX			UNIT
				MIN	TYP	MAX	
V <sub>IL</sub> *	Low-level input voltage					0.8	V
V <sub>IH</sub> *	High-level input voltage				2		V
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA		-0.8	-1.5	V
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V		-0.02	-0.25	mA
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V			25	μA
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V			1	mA
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 4mA	0.3	0.5	V
			COM	I <sub>OL</sub> = 4mA			
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -1mA	2.4	2.8	V
			COM	I <sub>OH</sub> = -1mA			
I <sub>OS</sub>	Output short-circuit current **	V <sub>CC</sub> = 5V	V <sub>O</sub> = 0V	-30	-70	-130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX			30	45	mA

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST	MILITARY			COMMERCIAL			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	R1 = 1.12kΩ R2 = 2.2kΩ		45	80		45	60	ns

Monolithic Memories

CUSTOM/SEMICUSTOM

**Half Power Series 20A-2**  
**16L8A-2, 16R8A-2, 16R6A-2, 16R4A-2**

**Operating Conditions**

SYMBOL	PARAMETER	MILITARY			COMMERCIAL			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V	
t <sub>w</sub>	Width of clock	Low	25	10	25	10		ns	
		High	25	10	25	10			
t <sub>su</sub>	Set up time from input or feedback to clock	16R6A-2	16R4A-2	16R8A-2	50	25	35	25	ns
t <sub>h</sub>	Hold time		0	-15	0	-15			ns
T <sub>A</sub>	Operating free-air temperature		-55		125		0	75	°C

**Electrical Characteristics Over Operating Conditions**

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
V <sub>IL</sub> *	Low-level input voltage					0.8	V	
V <sub>IH</sub> *	High-level input voltage			2			V	
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>i</sub> = -18mA		-0.8	-1.5	V	
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V		-0.02	-0.25	mA	
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V			25	μA	
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V			1	mA	
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 12mA	0.3	0.5	V	
			COM	I <sub>OL</sub> = 24mA				
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -2mA	2.4	2.8	V	
			COM	I <sub>OH</sub> = -3.2mA				
I <sub>OZL</sub>	Off-state output current †	V <sub>CC</sub> = MAX		V <sub>O</sub> = 0.4V			-100	μA
I <sub>OZH</sub>				V <sub>O</sub> = 2.4V			100	μA
I <sub>OS</sub>	Output short-circuit current **	V <sub>CC</sub> = 5V		V <sub>O</sub> = 0V	-30	-70	-130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX				60	90	mA

**Switching Characteristics Over Operating Conditions**

SYMBOL	PARAMETER		TEST CONDITIONS	MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	16L8A-2 16R6A-2 16R4A-2	R <sub>1</sub> = 200Ω R <sub>2</sub> = 390Ω		25	50	25	35	ns	
t <sub>CLK</sub>	Clock to output or feedback				15	25	15	25	ns	
t <sub>PXZ/ZX</sub>	Pin 11 to output disable/enable except 16L8A-2				15	25	15	25	ns	
t <sub>PZX</sub>	Input to output enable	16L8A-2 16R6A-2 16R4A-2			25	45	25	35	ns	
t <sub>PXZ</sub>	Input to output disable	16R8A-2 16R6A-2 16R4A-2			25	45	25	35	ns	
f <sub>MAX</sub>	Maximum frequency	16R8A-2 16R6A-2 16R4A-2			14	25	16	25	MHz	

## Operating Conditions

SYMBOL	PARAMETER			MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>CC</sub>	Supply voltage			4.5	5	5.5	4.75	5	5.25	V
t <sub>w</sub>	Width of clock	16R8A-4 16R6A-4 16R4A-4	Low	40	20		30	20		ns
			High	40	20		30	20		
t <sub>su</sub>	Set up time from input or feedback to clock	16R8A-4 16R6A-4 16R4A-4		90	45		60	45		ns
t <sub>h</sub>	Hold time			0	-15		0	-15		ns
T <sub>A</sub>	Operating free-air temperature					125			75	°C

## Electrical Characteristics Over Operating Conditions

SYMBOL	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
V <sub>IL</sub> *	Low-level input voltage					0.8	V	
V <sub>IH</sub> *	High-level input voltage			2			V	
V <sub>IC</sub>	Input clamp voltage	V <sub>CC</sub> = MIN	I <sub>I</sub> = -18mA		-0.8	-1.5	V	
I <sub>IL</sub>	Low-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 0.4V		-0.02	-0.25	mA	
I <sub>IH</sub>	High-level input current †	V <sub>CC</sub> = MAX	V <sub>I</sub> = 2.4V			25	μA	
I <sub>I</sub>	Maximum input current	V <sub>CC</sub> = MAX	V <sub>I</sub> = 5.5V			1	mA	
V <sub>OL</sub>	Low-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OL</sub> = 4mA	0.3	0.5	V	
			COM	I <sub>OL</sub> = 8mA				
V <sub>OH</sub>	High-level output voltage	V <sub>CC</sub> = MIN	MIL	I <sub>OH</sub> = -1mA	2.4	2.8	V	
			COM	I <sub>OH</sub> = -1mA				
I <sub>OZL</sub>	Output short-circuit current**	V <sub>CC</sub> = MAX		V <sub>O</sub> = 0.4V		-100	μA	
I <sub>OZH</sub>				V <sub>O</sub> = 2.4V		100	μA	
I <sub>OS</sub>	Output short-circuit current	V <sub>CC</sub> = 5V		V <sub>O</sub> = 0V	-30	-70	-130	mA
I <sub>CC</sub>	Supply current	V <sub>CC</sub> = MAX	16R4A-4 16R6A-4 16R8A-4 16L8A-4		30	50	mA	

## Switching Characteristics Over Operating Conditions

SYMBOL	PARAMETER		TEST	MILITARY			COMMERCIAL			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t <sub>PD</sub>	Input or feedback to output	16R6A-4 16R4A-4 16L8A-4	R <sub>1</sub> = 800Ω R <sub>2</sub> = 1.56kΩ		35	75		35	55	ns
t <sub>CLK</sub>	Clock to output or feedback				20	45		20	35	ns
t <sub>PXZ/ZX</sub>	Pin 11 to output disable/enable—except 16L8A-4				15	40		15	30	ns
t <sub>PZX</sub>	Input to output enable	16R6A-4 16R4A-4 16L8A-4			30	65		30	50	ns
t <sub>pxz</sub>	Input to output disable	16R6A-4 16R4A-4 16L8A-4			30	65		30	50	ns
f <sub>MAX</sub>	Maximum frequency	16R8A-4 16R6A-4 16R4A-4			8	18		11	18	MHz

2175 Mission College Boulevard, Santa Clara, CA 95050 Tel: (408) 970-9700 TWX: 910-338-2376 TWX: 910-338-2374

**Monolithic Memories** 

## Programming/Verifying Procedure

NOTES: For programming purposes many PAL pins have double functions.

### For The PAL 20:

As long as Pin 1 is at HH, Pin 11 is at ground, and Pin 12 is either at HH or Z (as defined in Table 1) — Pins 16, 17, 18, and 19 are outputs. The other pin functions are: I0 (Pin 2) through I7 (Pin 9) plus Pin 12 address the proper row; A0 (Pin 15), A1 (Pin 14), and A2 (Pin 13) address the proper product lines.

When Pin 11 is at HH, Pin 1 is at ground and Pin 19 is either at HH or Z — Pins 12, 13, 14, and 15 are outputs. The other pin functions are: I0 (Pin 2) through I7 (Pin 9) plus Pin 19 address the proper row; A0 (now Pin 18), A1 (now Pin 17), and A2 (now Pin 16) address the proper product lines.

### For The PAL 24:

As long as Pin 1 is at HH, Pin 13 is at ground and Pin 14 is either at HH or Z (as defined in Table 1) — Pins 19, 20, 21, 22, and 23 are outputs. The other pin functions are: I0 (Pin 2) through I9 (Pin 11) plus Pin 14 address the proper row; A0 (Pin 15), A1 (Pin 16), and A2 (Pin 17) address the proper product lines.

As long as Pin 13 is at HH, Pin 1 is at ground, and Pin 23 is either at HH or Z (as defined in Table 1) — Pins 14, 15, 16, 17, and 18 are outputs. The other pin functions are: I0 (Pin 2) through I9 (Pin 11) plus Pin 23 address the proper row; A0 (Pin 22), A1 (Pin 21), and A2 (Pin 20) address the proper product lines.

### For The PAL 24A:

As long as Pin 1 is at HH, Pin 13 is at ground, and Pin 14 is either at HH or Z (as defined in Table 1) — Pins 19, 20, 21, and 22 are outputs. The other pin functions are: I0 (Pin 2) through I9 (Pin 11) plus Pin 14 address the proper row; A0 (Pin 15), A1 (Pin 16), and A2 (Pin 17) address the proper product lines.

As long as Pin 13 is at HH, Pin 1 is at ground, and Pin 23 is either at HH or Z (as defined in Table 1) — pins 15, 16, 17, and 18 are outputs. The other Pin functions are: I0 (Pin 2) through I9 (Pin 11) plus Pin 23 address the proper row; A0 (Pin 22), A1 (Pin 21), and A2 (Pin 20) address the proper product lines.

## Pre-Verification

- 5.1.1 Raise  $V_{CC}$  to 5.0 volts.
- 5.1.2 Raise Output Disable pin, OD, to VIH.
- 5.1.3 Select an input line by specifying Inputs and L/R as shown in Table 1 or Table 2.
- 5.1.4 Select a product line by specifying A0, A1, and A2 one-of-eight select as shown in Table 3, Table 4 or Table 5.
- 5.1.5 Pulse the CLOCK pin and verify (with CLOCK at VIL) that the output pin, O, is in the state corresponding to an unblown fuse.
  - For verified unblown condition, continue procedure from 5.1.3 through 5.1.5.
  - For verified blown condition, stop procedure and reject part.

## Programming Algorithm

- 5.2.1 Raise Output Disable pin, OD, to VIH.
- 5.2.2 Programming pass. For all fuses to be blown:
  - 5.2.2.1 Lower CLOCK pin to ground.
  - 5.2.2.2 Select an input line by specifying Inputs and L/R as shown in Table 1 or Table 2.
  - 5.2.2.3 Select a product line by specifying A0, A1, and A2 one-of-eight select as shown in Table 3, Table 4 or Table 5.
  - 5.2.2.4 Raise  $V_{CC}$  to VIH.
  - 5.2.2.5 Program the fuse by pulsing the output pins of the selected product group -one at a time- to VIH (as shown in the Programming Waveforms, Section 5.5).
  - 5.2.2.6 Lower  $V_{CC}$  to 5.0 volts.
  - 5.2.2.7 Repeat this procedure from 5.2.2.2 until pattern is complete.
- 5.2.3 First verification pass. For all fuse locations:
  - 5.2.3.1 Select an input line by specifying Inputs and L/R as shown in Table 1 or Table 2.
  - 5.2.3.2 Select a product line by specifying A0, A1, and A2 one-of-eight select as shown in Table 3, Table 4 or Table 5.
  - 5.2.3.3 Pulse the CLOCK pin and verify (with CLOCK at VIL) that the output pin, O, is in the correct state.
    - For verified output state, continue procedure
    - For overblow condition, stop procedure and reject part.
    - For underblow condition, reexecute steps 5.2.2.4 through 5.2.2.6 and 5.2.2.3. If successful, continue procedure. After three attempts to blow fuse without success, reject part but continue procedure.
  - 5.2.3.4 Repeat this procedure from 5.2.3.1 until the entire array is exercised.
- 5.2.4 High Voltage Verify. For all fuse locations:
  - 5.2.4.1 Raise  $V_{CC}$  to 5.5 volts.
  - 5.2.4.2 Select an input line by specifying Inputs and L/R as shown in Table 1 or Table 2.
  - 5.2.4.3 Select a product line by specifying A0, A1, and A2 one-of-eight select as shown in Table 3, Table 4 or Table 5.
  - 5.2.4.4 Pulse the CLOCK pin and verify (with CLOCK at VIL) that the output pin, O, is in the correct state.
    - For verified output state, continue procedure
    - For invalid output state, stop procedure and reject part.
  - 5.2.4.5 Repeat this procedure from 5.2.4.1 until the entire array is exercised.

5.2.5 Low Voltage Verify. For all fuse locations:

- 5.2.5.1 Lower  $V_{CC}$  to 4.5 volts.
- 5.2.5.2 Select an input line by specifying Inputs and L/R as shown in Table 1 or Table 2.
- 5.2.5.3 Select a product line by specifying A0, A1, and A2, one-of-eight select as shown in Table 3, Table 4 or Table 5.
- 5.2.5.4 Pulse the CLOCK pin and verify (with CLOCK at VIL) that the output pin, O, is in the correct state.
  - For verified output state, continue procedure.
  - For invalid output state, continue procedure and reject part.

## Programming the Security Fuses

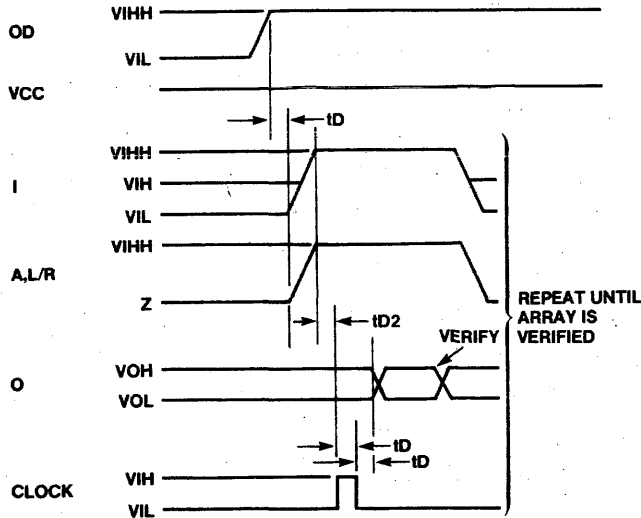
- 5.3.1 Verify per Section 5.2.4 and 5.2.5.
- 5.3.2 Raise  $V_{CC}$  to 6 volts.
- 5.3.3 For PAL 20:
  - Program the first fuse by pulsing Pin 1 to VP. (From 1 to 5 pulses is acceptable.)
  - Program the second fuse by pulsing Pin 11 to VP. (From 1 to 5 pulses is acceptable.)
- 5.3.4 For PAL 24 and PAL 24A:
  - Program the first fuse by pulsing Pin 1 to VP. (From 1 to 5 pulses is acceptable.)
  - Program the second fuse by pulsing Pin 13 to VP. (From 1 to 5 pulses is acceptable.)
- 5.3.5 Verify per Section 5.2.4 and 5.2.5:
  - A device is "secure" if either half fails to verify.

## 5.4 Programming Parameters

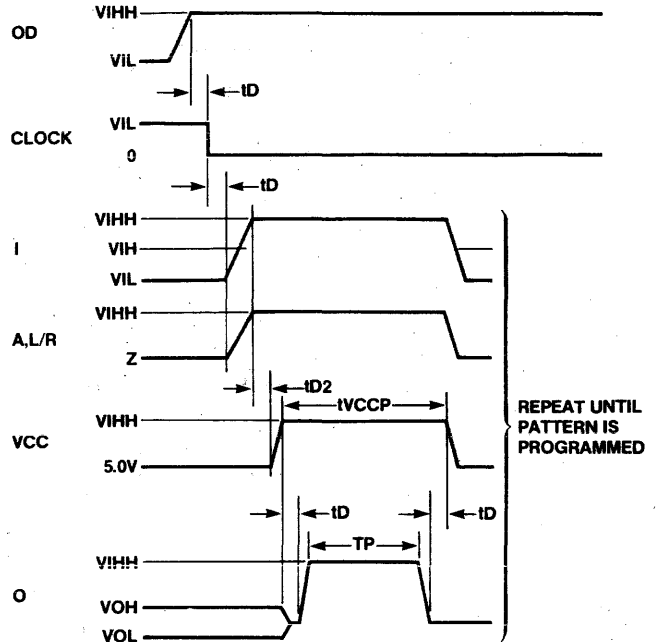
SYMBOL	PARAMETER	LIMITS			UNIT
		MIN	TYP	MAX	
$V_{IHH}$	Program-level input voltage	11.5	11.75	12	V
$I_{IHH}$	Output Program Pulse			50	mA
	OD, L/R			50	
	All other inputs			10	
$I_{CCH}$	Program Supply Current			900	mA
$t_{VCCP}$	Pulse Width of $V_{CC}$ @ $V_{IHH}$			60	$\mu$ S
$T_P$	Program Pulse Width	10	20	50	$\mu$ S
$t_D$	Delay Time	100			ns
$t_{D2}$	Delay Time after L/R Pin	10			$\mu$ S
	$V_{CCP}$ Duty Cycle			20	%
$V_P$	Security Fuse Programming Voltage	18	18.5	19	V
$I_P$	Security Fuse Programming Supply Current			400	mA
$T_{PP}$	Security Fuse Programming Pulse Width	10	40	70	$\mu$ S
	Security Fuse Programming Duty Cycle			50	%
$t_{RP}$	Rise time of output programming and address pulses	1	1.5	10	V/ $\mu$ S
$t_{RP}$	Rise Time of security fuse programming pulses	1	1.5	10	V/ $\mu$ S
$V_{CCPP}$	$V_{CC}$ value during security fuse programming	5.75	6.0	6.25	V
	$V_{CC}$ value for first verify	4.75	5.0	5.25	
	$V_{CC}$ value for High $V_{CC}$ verify	5.4	5.5	5.6	
	$V_{CC}$ value for Low $V_{CC}$ verify	4.4	4.5	4.6	

# PAL Programming

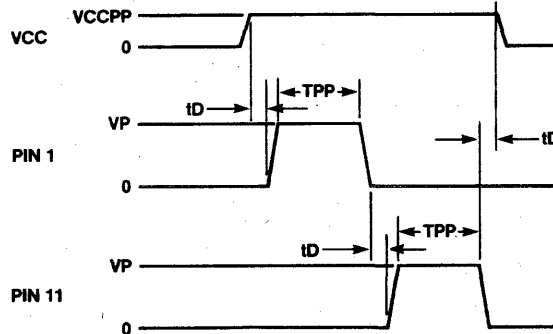
## 5.5 ARRAY PROGRAMMING



NOTE: VCC (Low Voltage Verify) = 4.5 volts  
 VCC (High Voltage Verify) = 5.5 volts  
 VCC (First Verify) = 5.0 volts  
 A Delay ( $t_{D2}$ ) must always precede the Positive Clock Transition. (e.g see section 5.2.3.3 for underblow condition)



## 5.6 SECURITY FUSE PROGRAMMING



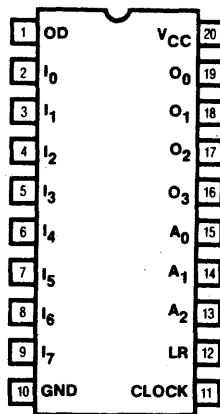
## Programmer/Development Systems

VENDOR	PAL 20s (ALL)	PAL 24s (STD)	PAL 24s (FAST)
Data I/O	— LogicPak (Rev-010) — 1427 Card Set	LogicPak (Rev-010)	— LogicPak (Rev-010)
Structured Design	— SD 20/24 — PAL Burner *	— SD 20/24 — PAL Burner *	— SD 20/24 — PAL Burner *
STAG	— PM202 (Rev 3) — PM2200 *	— PM202 (Rev 3) — PM2200 *	— PM202 (Rev) — PM2200 *
DIGELEC	— UP803 (FAM51) or (FAM52)	— UP803 (FAM51) or (FAM52)	— UP803 (FAM51) or (FAM52)
PROLOG	— M980 PM9068		
KONTRON	— MPP80S MOD 21		

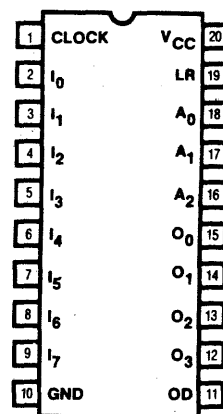
\*Means that this version is being qualified.

Programming Pin Configurations

PRODUCTS 0 THRU 31



PRODUCTS 32 THRU 63



Voltage Legend

L = Low-level input voltage,  $V_{IL}$   
 H = High-level input voltage,  $V_{IH}$

HH = High-level program voltage,  $V_{IHH}$   
 Z = High impedance (e.g., 10k $\Omega$  to 5.0V)

INPUT LINE NUMBER	PIN IDENTIFICATION								L/R
	I7	I6	I5	I4	I3	I2	I1	I0	
0	HH	HH	HH	HH	HH	HH	HH	L	Z
1	HH	HH	HH	HH	HH	HH	HH	H	Z
2	HH	HH	HH	HH	HH	HH	HH	L	HH
3	HH	HH	HH	HH	HH	HH	HH	H	HH
4	HH	HH	HH	HH	HH	HH	L	HH	Z
5	HH	HH	HH	HH	HH	HH	H	HH	Z
6	HH	HH	HH	HH	HH	HH	L	HH	HH
7	HH	HH	HH	HH	HH	HH	H	HH	HH
8	HH	HH	HH	HH	HH	L	HH	HH	Z
9	HH	HH	HH	HH	HH	H	HH	HH	Z
10	HH	HH	HH	HH	HH	L	HH	HH	HH
11	HH	HH	HH	HH	HH	H	HH	HH	HH
12	HH	HH	HH	HH	L	HH	HH	HH	Z
13	HH	HH	HH	HH	H	HH	HH	HH	Z
14	HH	HH	HH	HH	L	HH	HH	HH	HH
15	HH	HH	HH	HH	H	HH	HH	HH	HH
16	HH	HH	HH	L	HH	HH	HH	HH	Z
17	HH	HH	HH	H	HH	HH	HH	HH	Z
18	HH	HH	HH	L	HH	HH	HH	HH	HH
19	HH	HH	HH	H	HH	HH	HH	HH	HH
20	HH	HH	L	HH	HH	HH	HH	HH	Z
21	HH	HH	H	HH	HH	HH	HH	HH	Z
22	HH	HH	L	HH	HH	HH	HH	HH	HH
23	HH	HH	H	HH	HH	HH	HH	HH	HH
24	HH	L	HH	HH	HH	HH	HH	HH	Z
25	HH	H	HH	HH	HH	HH	HH	HH	Z
26	HH	L	HH	HH	HH	HH	HH	HH	HH
27	HH	H	HH	HH	HH	HH	HH	HH	HH
28	L	HH	HH	HH	HH	HH	HH	HH	Z
29	H	HH	HH	HH	HH	HH	HH	HH	Z
30	L	HH	HH	HH	HH	HH	HH	HH	HH
31	H	HH	HH	HH	HH	HH	HH	HH	HH

Table 1 Input Line Select

PRODUCT LINE NUMBER	PIN IDENTIFICATION						
	O3	O2	O1	O0	A2	A1	A0
0, 32	Z	Z	Z	HH	Z	Z	Z
1, 33	Z	Z	Z	HH	Z	Z	HH
2, 34	Z	Z	Z	HH	Z	HH	Z
3, 35	Z	Z	Z	HH	Z	HH	HH
4, 36	Z	Z	Z	HH	HH	Z	Z
5, 37	Z	Z	Z	HH	HH	Z	HH
6, 38	Z	Z	Z	HH	HH	HH	Z
7, 39	Z	Z	Z	HH	HH	HH	HH
8, 40	Z	Z	HH	Z	Z	Z	Z
9, 41	Z	Z	HH	Z	Z	Z	HH
10, 42	Z	Z	HH	Z	Z	HH	Z
11, 43	Z	Z	HH	Z	Z	HH	HH
12, 44	Z	Z	HH	Z	HH	Z	Z
13, 45	Z	Z	HH	Z	HH	Z	HH
14, 46	Z	Z	HH	Z	HH	HH	Z
15, 47	Z	Z	HH	Z	HH	HH	HH
16, 48	Z	HH	Z	Z	Z	Z	Z
17, 49	Z	HH	Z	Z	Z	Z	HH
18, 50	Z	HH	Z	Z	Z	HH	Z
19, 51	Z	HH	Z	Z	Z	HH	HH
20, 52	Z	HH	Z	Z	HH	Z	Z
21, 53	Z	HH	Z	Z	HH	Z	HH
22, 54	Z	HH	Z	Z	HH	HH	Z
23, 55	Z	HH	Z	Z	HH	HH	HH
24, 56	HH	Z	Z	Z	Z	Z	Z
25, 57	HH	Z	Z	Z	Z	Z	HH
26, 58	HH	Z	Z	Z	Z	HH	Z
27, 59	HH	Z	Z	Z	Z	HH	HH
28, 60	HH	Z	Z	Z	HH	Z	Z
29, 61	HH	Z	Z	Z	HH	Z	HH
30, 62	HH	Z	Z	Z	HH	HH	Z
31, 63	HH	Z	Z	Z	HH	HH	HH

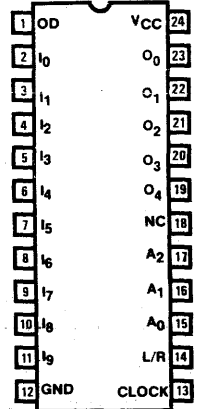
Table 2 Product Line Select

Monolithic Memories

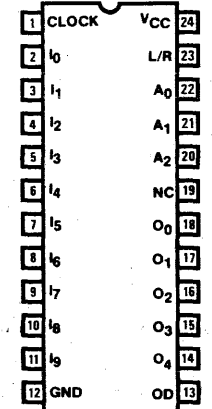
CUSTOM/SEMICUSTOM

Programming Pin Configurations

PRODUCTS 0 THRU 39



PRODUCTS 40 THRU 79



Voltage Legend

L = Low-level input voltage,  $V_{IL}$   
 H = High-level input voltage,  $V_{IH}$

HH = High-level program voltage,  $V_{IHH}$   
 Z = High impedance (e.g. 10K  $\Omega$  to 5.0V)

INPUT LINE NUMBER	PIN IDENTIFICATION										
	I <sub>9</sub>	I <sub>8</sub>	I <sub>7</sub>	I <sub>6</sub>	I <sub>5</sub>	I <sub>4</sub>	I <sub>3</sub>	I <sub>2</sub>	I <sub>1</sub>	I <sub>0</sub>	L/R
0	HH	HH	HH	HH	HH	HH	HH	HH	HH	L	Z
1	HH	HH	HH	HH	HH	HH	HH	HH	HH	H	Z
2	HH	HH	HH	HH	HH	HH	HH	HH	HH	L	HH
3	HH	HH	HH	HH	HH	HH	HH	HH	HH	H	HH
4	HH	HH	HH	HH	HH	HH	HH	HH	L	HH	Z
5	HH	HH	HH	HH	HH	HH	HH	HH	H	HH	Z
6	HH	HH	HH	HH	HH	HH	HH	HH	L	HH	HH
7	HH	HH	HH	HH	HH	HH	HH	HH	H	HH	HH
8	HH	HH	HH	HH	HH	HH	HH	L	HH	HH	Z
9	HH	HH	HH	HH	HH	HH	HH	H	HH	HH	Z
10	HH	HH	HH	HH	HH	HH	HH	L	HH	HH	HH
11	HH	HH	HH	HH	HH	HH	HH	H	HH	HH	HH
12	HH	HH	HH	HH	HH	HH	L	HH	HH	HH	Z
13	HH	HH	HH	HH	HH	HH	H	HH	HH	HH	Z
14	HH	HH	HH	HH	HH	HH	L	HH	HH	HH	HH
15	HH	HH	HH	HH	HH	HH	H	HH	HH	HH	HH
16	HH	HH	HH	HH	HH	L	HH	HH	HH	HH	Z
17	HH	HH	HH	HH	HH	H	HH	HH	HH	HH	Z
18	HH	HH	HH	HH	HH	L	HH	HH	HH	HH	HH
19	HH	HH	HH	HH	HH	H	HH	HH	HH	HH	HH
20	HH	HH	HH	HH	L	HH	HH	HH	HH	HH	Z
21	HH	HH	HH	HH	H	HH	HH	HH	HH	HH	Z
22	HH	HH	HH	HH	L	HH	HH	HH	HH	HH	HH
23	HH	HH	HH	HH	H	HH	HH	HH	HH	HH	HH
24	HH	HH	HH	L	HH	HH	HH	HH	HH	HH	Z
25	HH	HH	HH	H	HH	HH	HH	HH	HH	HH	Z
26	HH	HH	HH	L	HH	HH	HH	HH	HH	HH	HH
27	HH	HH	HH	H	HH	HH	HH	HH	HH	HH	HH
28	HH	HH	L	HH	HH	HH	HH	HH	HH	HH	Z
29	HH	HH	H	HH	HH	HH	HH	HH	HH	HH	Z
30	HH	HH	L	HH	HH	HH	HH	HH	HH	HH	HH
31	HH	HH	H	HH	HH	HH	HH	HH	HH	HH	HH
32	HH	L	HH	HH	HH	HH	HH	HH	HH	HH	Z
33	HH	H	HH	HH	HH	HH	HH	HH	HH	HH	Z
34	HH	L	HH	HH	HH	HH	HH	HH	HH	HH	HH
35	HH	H	HH	HH	HH	HH	HH	HH	HH	HH	HH
36	L	HH	HH	HH	HH	HH	HH	HH	HH	HH	Z
37	H	HH	HH	HH	HH	HH	HH	HH	HH	HH	Z
38	L	HH	HH	HH	HH	HH	HH	HH	HH	HH	HH
39	H	HH	HH	HH	HH	HH	HH	HH	HH	HH	HH

Table 1 Input Line Select

PRODUCT LINE NUMBER	PIN IDENTIFICATION							
	O <sub>4</sub>	O <sub>3</sub>	O <sub>2</sub>	O <sub>1</sub>	O <sub>0</sub>	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>
0, 40	Z	Z	Z	Z	HH	Z	Z	Z
1, 41	Z	Z	Z	Z	HH	Z	Z	HH
2, 42	Z	Z	Z	Z	HH	Z	HH	Z
3, 43	Z	Z	Z	Z	HH	Z	HH	HH
4, 44	Z	Z	Z	Z	HH	HH	Z	Z
5, 45	Z	Z	Z	Z	HH	HH	Z	HH
6, 46	Z	Z	Z	Z	HH	HH	HH	Z
7, 47	Z	Z	Z	Z	HH	HH	HH	HH
8, 48	Z	Z	Z	HH	Z	Z	Z	Z
9, 49	Z	Z	Z	HH	Z	Z	Z	HH
10, 50	Z	Z	Z	HH	Z	Z	HH	Z
11, 51	Z	Z	Z	HH	Z	Z	HH	HH
12, 52	Z	Z	Z	HH	Z	HH	Z	Z
13, 53	Z	Z	Z	HH	Z	HH	Z	HH
14, 54	Z	Z	Z	HH	Z	HH	HH	Z
15, 55	Z	Z	Z	HH	Z	HH	HH	HH
16, 56	Z	Z	HH	Z	Z	Z	Z	Z
17, 57	Z	Z	HH	Z	Z	Z	Z	HH
18, 58	Z	Z	HH	Z	Z	Z	HH	Z
19, 59	Z	Z	HH	Z	Z	Z	HH	HH
20, 60	Z	Z	HH	Z	Z	HH	Z	Z
21, 61	Z	Z	HH	Z	Z	HH	Z	HH
22, 62	Z	Z	HH	Z	Z	HH	HH	Z
23, 63	Z	Z	HH	Z	Z	HH	HH	HH
24, 64	Z	HH	Z	Z	Z	Z	Z	Z
25, 65	Z	HH	Z	Z	Z	Z	Z	HH
26, 66	Z	HH	Z	Z	Z	Z	HH	Z
27, 67	Z	HH	Z	Z	Z	Z	HH	HH
28, 68	Z	HH	Z	Z	Z	HH	Z	Z
29, 69	Z	HH	Z	Z	Z	HH	Z	HH
30, 70	Z	HH	Z	Z	Z	HH	HH	Z
31, 71	Z	HH	Z	Z	Z	HH	HH	HH
32, 72	HH	Z	Z	Z	Z	Z	Z	Z
33, 73	HH	Z	Z	Z	Z	Z	Z	HH
34, 74	HH	Z	Z	Z	Z	Z	HH	Z
35, 75	HH	Z	Z	Z	Z	Z	HH	HH
36, 76	HH	Z	Z	Z	Z	HH	Z	Z
37, 77	HH	Z	Z	Z	Z	HH	Z	HH
38, 78	HH	Z	Z	Z	Z	HH	HH	Z
39, 79	HH	Z	Z	Z	Z	HH	HH	HH

Table 2 Product Line Select



# FOR CUSTOM LSI/VLSI BIPOLAR CIRCUITS

## High Performance ALS-TTL-Compatible Macrocell Arrays

In addition to standard logic lines, Motorola also offers a variety of TTL-compatible Macrocell Arrays. These products provide a means for developing economical custom LSI/VLSI logic circuits. Performance is achieved by the combination of an advanced MOSAIC I (Motorola Oxide-Isolated Self-Aligned Implanted Circuit) oxide isolated bipolar integrated circuit process and a series gated emitter-coupled logic (ECL) macrocell circuit technology. Input and output circuits provide level translation to and from the internal array logic for standard TTL/MOS interface.

Each cell within the arrays contains a number of unconnected transistors and resistors. Stored within a computer are the specifications to automatically interconnect these elements forming SSI/MSI logic cells (rather than simple gates) called macrocells. These macrocells take the form of standard logic blocks such as dual type D flip-flops, dual full adders, quad latches and many other pre-defined "library" functions. Presently, the macrocell library for the ALS-TTL arrays contains more than 80 logic functions.

Generating an LSI/VLSI design is simply a matter of selecting the appropriate macrocells and describing the proper interconnection network to implement the design. Motorola's CAD (Computer-Aided-Design) interface provides automatic placement and routing of the cells (intraconnection of the cell itself is automatically accom-

## MCA500ALS MCA1300ALS MCA2800ALS

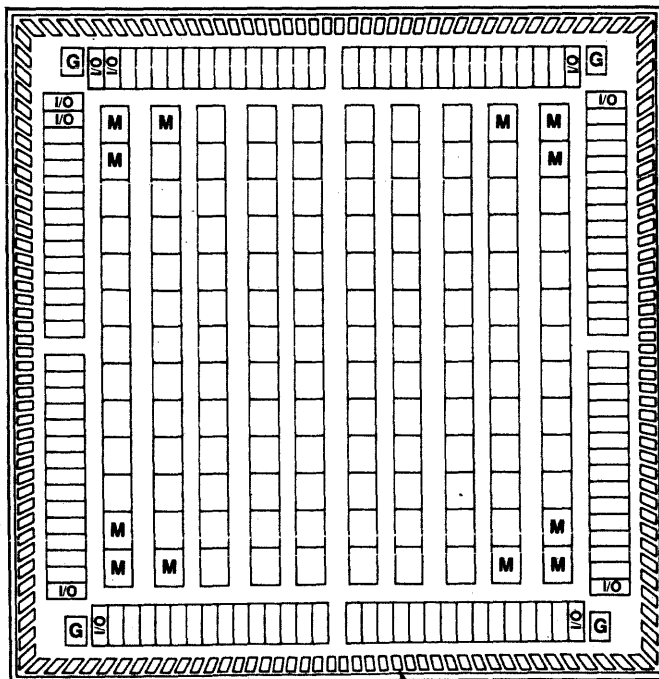
plished when placed), full logic and fault-testing capabilities, AC delay simulations, generation of test tapes and custom metallization to complete the IC processing sequence.

The ability to stockpile fully diffused wafers provides a very fast turnaround time (the time from customer notification of a completed design to delivery of finished parts) of currently nine weeks.

### ALS-TTL MCA Selection Chart

	MCA 500ALS	MCA 1300ALS	MCA 2800ALS	MCA 2400ETL
Configuration				
Max Gate Equivalent	533	1280	2720	2958
Major Macrocells	24	60	30	130
I/O Ports	57	76	120	120
Input/Interface Cells	26	40	120 I/O	70TTL I/O
Output Macrocells	24	40		72ECL
Performance				
Max Gate Delay (M Cell)	4.0 ns	3.0 ns	1.1 ns	1.1 ns
Max Toggle Frequency	80 MHz	80 MHz	125 MHz	125 MHz
Maximum Power Dissipation	1.0 W	1.4 W	2.5 W	4.0 W
Packages				
Dual-In-Line	28, 40, 48	40, 48	—	—
Chip Carrier, Pin Grid Array	68	68, 84	84, 149PG	149PG
Temperature Range	0-70°C	0-70°C	0-70°C	0-70°C
Supply Voltage	5.0 V ±5%	5.0 V ±5%	5.0 V ±5%	+5.0, -5.2 V
Availability	Now	Now	Now	4Q83

### MCA2800ALS MACROCELL ARRAY LAYOUT



- M** — Major (Internal) Cells  
Divisible to Four ¼ Cells  
130 Total
- I/O** — Input/Output Cells  
120 Total
- G** — Master Bias Generators

168 Bonding Pads

# MECL 10,000 Macrocell Array Family

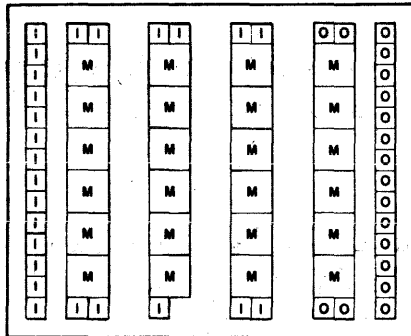


The Macrocell Array Concept combines a pre-diffused array of components with computer-aided design techniques to offer system designers a rapid, cost-effective means for implementing semi-custom, high-speed digital logic systems with VLSI circuitry.

Compared with equivalent systems developed with discrete logic (separately packed SSI/MSI logic functions) the high packing density of the Macrocell array chip offers up to 100-to-1 reduction in system component count, with a power dissipation improvement (reduction) of as much as 12 to 1.

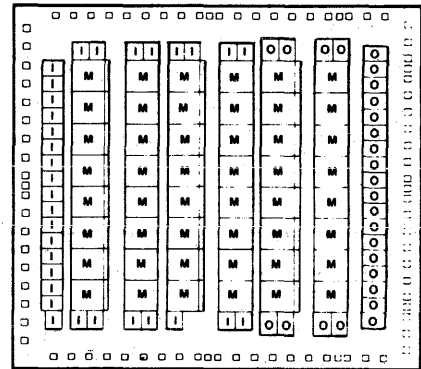
Compared with the conventional approach to custom LSI circuits, the Macrocell approach offers a tremendous reduction in delivery time. With a stockpile of fully diffused wafers, turnaround time (from the time the customer gives the go-ahead signal for generating masks until he receives finished parts) is seven weeks.

MECL Macrocell Array family members presently consist of the MCA1200ECL (68 Pin) Array, the MCA600ECL (28, 40 and 68 Pin) Array, and the MCA2500ECL (149 Pin) Array.



**THE MCA600ECL**

Total of 67 cell locations—24 Major, 25 Interface and 18 Output cell locations.



**THE MCA1200ECL**

Total of 106 cell locations—48 Major, 32 Interface and 26 Output cell locations.

Parameter	MCA2500ECL	MCA1200ECL	MCA600ECL
Maximum Equivalent Gates	2472	1192	652
Major Macrocells	110	48	24
Interface/Input Macrocells	—	32	25
Output Macrocells	68	26	18
Maximum Gate Delay	0.5 ns	1.2 ns	1.2 ns
Power Dissipation	6.0 Watts	4 Watts	2.5 Watts
Package	149 Pin	68 Leadless	28,40 DIL
	Grid Array	72 Pin Grid	68 Leadless
Temperature Range	0–70°C	0–70°C	0–70°C
I/O Interface	10K/10KH/100K	MECL 10K/10KH	MECL 10K/10KH
Design Interface	CAD	CAD	CAD

## M10900 Family

The M10900 family is a series of very high performance bipolar LSI products designed off the MECL Macrocell Array offering. While the Macrocell Array is normally used for

custom circuits which require option development time and costs, the M10900 family is a standard product ordered like any other MECL 10K circuit.

Device	Description	Comments
MC10900Z	8-Bit ALU with Parity Slice	17.8 ns Max Addition of 2 8-Bit Words. Introduced
MC10901Z	8 x 8-Bit Array Multiplier	Expandable to 32 x 32. 24.3 ns Max Delay for 8 x 8 Multiply. Introduced
MC10902Z	8-Bit Binary/BCD ALU Slice	Expandable to 32-Bits. 17.9 ns Max Add Delay. Introduced
MC10904Z	8-Bit Micro Code Sequencer Slice	Expandable to 24-Bits. 14.0 ns Max Clock to Output.
MC10905Z	Error Detect and Correct Circuit	16-Bit Check and Correct with 1 Chip. Expandable to 64-Bits. 16-Bit Check and Correct in 19.7 ns Max.

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**Description**

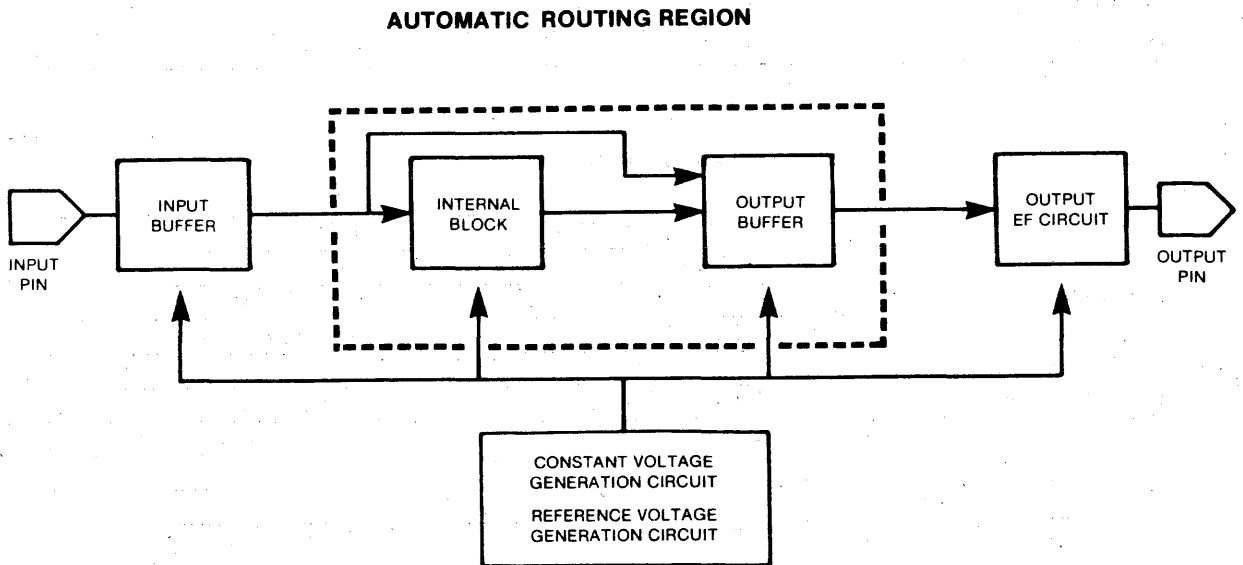
The μPB6300 series consists of three high speed ECL gate arrays manufactured using advanced bipolar technology. The μPB6301, μPB6310, and μPB6320 have 300, 1200, and 2000 equivalent gates, respectively. Specific LSI designs are built by interconnecting cells (2.5 gates) by means of advanced CAD tools such as logic simulation, automatic placement and routing, delay simulation, and test program generation. Customers can use the GE MARK III® time sharing system through their own terminals for logic simulation.

Customers can choose from a wide range of package types. The μPB6300 gate arrays are available in 72 = to 132 = pin grid arrays (PGA) packages and a 68-pin leadless dip carrier (LCC) package.

**Features**

- High speed:  $\square$  0.5ns/gate (fan-in (=) fan-out (=) 2.5, L = 2.0mm)
- ECL 100K compatible
- Shallow junction dielectric isolation
- 68 available macros in function block library
- Fully supported by advanced CAD with direct customer access
- Quick Turn Around Time  $\square$  available 8~10 weeks after logic validation

**CONSTITUTION**



BLOCK	CONSTITUTION	FUNCTION	BLOCK HEADING
Input, Output Pin		ECL -100 K level interface	—
Input Buffer	EF output, CML internal	Level shift ECL - 100 K to CML	M
Internal Block	EF Output, CML internal	Shown in block library	C, J, V
Output Buffer	ECL -100 K Basic Circuit (W/O output EF)	level shift CML to ECL - 100 K (With simple logic function)	G
Output Special Circuit	only EF for Output	For ECL -100 k level output	B
Constant Voltage	Constant voltage	1) C.S. constant current control	—
Generation Reference Voltage Generation	Generation Circuit with Temperature compensation	2) Generation of reference voltage	—

**μPB6301 SPECIFICATION**

ITEM	SPECIFICATION		COMMENT	
Part Number	μPB6301			
Integration (Gate)	~ 300		1 cell = 2.25 gate	
Input Output Interface	ECL - 100 K		output: 50 Ω to -2 V	
Power Supply Voltage	VEE = -4.5 V ± 10%		max = -5.7 V	
Ambient Temperature	Ta = Q ~ +70°C		4 m/sec air cooling	
t <sub>pd</sub> /P <sub>o</sub>	Internal Gate	0.5 nS/5.4 mW	F/I = F/O = 2.5, L = 2 mm	
	Input Buffer	0.5 nS/5.4 mW		
	Output Buffer	0.8 nS/22.5 mW		
Power Dissipation	3.1 W (typ)		Ta, VEE = Typical	
	4.6 W (max)		Ta, VEE = Worst	
Package	72 PIP	68 LCC	θ = 5.6°C/W,	
Pin Count (Max.)	Input	56	56	0 = 10°C/W, 1m/sec.
	Output	28	28	Input + Output ≤ 56
	Power Supply	8	8	
Cell	Internal cell	16 × 8 = 128		
Count	Peripheral	14 × 2 = 28		
Assembly Method	Wire bonding			
Number of Macros	68			
Circuit Form	EF Output, CML Internal		with wired logic	

**μPB6310 SPECIFICATION**

ITEM	SPECIFICATION			COMMENT	
Part Number	μPB6310				
Integration (Gate)	~ 1200			1 cell = 2.25 gate	
Input Output	ECL - 100 K			outputs: 50 Ω to -2 V	
Power Supply Voltage	VEE = -4.5 V ± 10%			max = -5.7 V	
Ambient Temperature	Ta = O ~ +70°C			4 m/sec air cooling	
t <sub>pd</sub> /P <sub>o</sub>	Internal	0.7 nS/1.9 mW		F/I = F/o = 2.5, L = 3 mm	
	Input Buffer	0.6 nS/1.9 mW			
	Output Buffer	1.0 nS/22.5 mW			
Power Dissipation	3.7 W (typ)			Ta, VEE = Typical	
	5.6 W (max)			Ta, VEE = Worst	
Package	72 PIP	132 PIP	68 LCC	θ = 5.6°C/W,	
Pin Count (Max.)	Input	64	88	60	Input + Output ≤
	Output	32	48	32	Package Pin - Power Pin
	Power Supply	8	16	8	
Cell	Internal	32 × 16 = 512			
Count	Peripheral	24 × 2 = 48			
Assembly Method	Wire bonding				
Number of Macros	50				
Circuit Form	EF Output, CML Internal			with wired logic	

**μPB6320 SPECIFICATION**

ITEM	SPECIFICATION	COMMENT	
Part Number	μPB6320		
Integration (Gate)	~ 2000	1 cell = 2.25 gate	
Input Output Interface	ECL - 100 K	output: 50 Ω to -2.0 V	
Power Supply Voltage	V <sub>EE</sub> = -4.5 V ± 10%	max = -5.7 V	
Ambient Temperature	T <sub>a</sub> = 0 ~ +70°C	4 m/sec air cooling	
t <sub>pd</sub> /P <sub>o</sub>	Internal Gate	0.7 nS/1.9 mW	
	Input Buffer	0.6 nS/1.9 mW	
	Output Buffer	1.0 nS/22.5 mW	
Power Dissipation	6.1 W (typ)	T <sub>a</sub> , V <sub>EE</sub> = Typical	
	9.5 W (max)	T <sub>a</sub> , V <sub>EE</sub> = Worst	
Package	72 PIP 132 PIP 68 LCC	θ = 5.6° C/W,	
Pin Count (Max.)	Input	56 108 56	Used Input + Used
	Output	28 48 28	Output ≤ Max Input
	Power Supply	16 24 16	
Cell	Internal cell	32 × 26 = 832	
Count	Peripheral	24 × 2 = 48	
Assembly Method	Wire bonding		
Number of Macros	50		
Circuit Form	EF Output, CML Internal	with wired logic	

μPB6301, μPB6310, μPB6320

Function Block Library List

INPUT BUFFER

Block Name	Function	6301 Typical Power (mW)	Cell Count	6310/6320 Typical Power (mW)	Cell Count
M201	1-IN-AND	11.5	1	3.7	1
M210	1-IN-NAND	11.5	1	3.7	1
M211	1-IN-AND (low power)	8.6	1	1.9	1
M212	1-IN-NAND (low power)	8.6	1	1.9	1

INTERNAL BLOCK

Block Name	Function	6301 Typical Power (mW)	Cell Count	6310/6320 Typical Power (mW)	Cell Count
JW1	2-IN-WIRED-AND	-5.9	0	-1.8	0
JW13	3-IN-WIRED-AND	-13.2	0	-2.7	0
JW14	4-IN-WIRED-AND	-17.7	0	-5.6	0
V001	CLAMP-VOLTAGE-GENERATOR	35.8	1	9.8	1
C027	1-IN 2-OUTPUT AND/NAND	29.2	1	9.1	1
C028	2-IN 2-OUTPUT AND/NAND	29.2	1	9.1	1
C029	3-IN 2-OUTPUT AND/NAND	29.2	1	9.1	1
C030	4-IN 2-OUTPUT AND/NAND	29.2	1	9.1	1
C031	5-IN 2-OUTPUT AND/NAND	29.2	1	9.1	1
C004	DUAL 1-IN-AND/NAND	34.6	1	11.0	1
C005	DUAL 2-IN-AND/NAND	34.6	1	11.0	1
C032	1-IN 3-OUTPUT AND 1-OUTPUT NAND BUFFER	29.2	1	9.1	1
C033	1-IN 1-OUTPUT AND 3-OUTPUT NAND BUFFER	29.2	1	9.1	1
C034	2-IN 3-OUTPUT AND 1-OUTPUT NAND BUFFER	29.2	1	9.1	1
C035	2-IN 1-OUTPUT AND 3-OUTPUT NAND BUFFER	29.2	1	9.1	1
C036	3-IN 3-OUTPUT AND 1-OUTPUT NAND BUFFER	29.2	1	9.1	1
C037	3-IN 1-OUTPUT AND 3-OUTPUT NAND BUFFER	29.2	1	9.1	1

C038	1-IN-AND-SIN-BUFFER	16.7	3	—	—
C039	1-IN-NAND-DELAY-BUFFER	28.1	4	—	—
C040	DUAL-1-IN-DELAY-BUFFER	16.7	1	—	—
C041	2-WIDE 2·2-IN-AND-OR	34.6	1	—	—
C042	3-WIDE 3·2·2-IN-AND-OR	63.7	3	—	—
C043	4-WIDE 3·3·3·3-IN-AND-OR	81.0	3	—	—
C044	2-WIDE 3·3-IN-AND-OR	46.4	2	—	—
C045	2-WIDE 1·1-IN-AND-OR	34.6	1	11.0	1
C046	2-WIDE 1·1-IN-AND-OR	34.6	1	11.0	1
C047	2-WIDE 1·2-IN-AND-OR	34.6	1	11.0	1
C048	3-WIDE 1·2·3-IN-AND-OR	45.9	2	14.7	2
C049	4-WIDE 1·2·3·4-IN-AND-OR	57.2	3	18.4	3
C050	5-WIDE 1·2·3·4·4-IN-AND-OR	68.6	3	—	—
C051	2-WIDE 2·2-IN-AND-OR	28.6	1	9.2	1
C052	2-WIDE 2·2-IN-AND-OR	46.4	2	14.6	2
C053	2-WIDE 3·3-IN-AND-OR	46.4	2	24.2	2
C054	2-WIDE 4·4-IN-AND-OR	46.4	2	24.2	2
C055	3-WIDE 2·2·2-IN-AND-OR	63.7	3	33.3	3
C056	3-WIDE 3·3·3-IN-AND-OR	45.9	2	24.2	2
C057	4-WIDE 2·2·2·2-IN-AND-OR	51.3	2	27.1	2
C058	4-WIDE 2·2·2·2-IN-AND-OR	39.4	2	13.0	2
C059	1·1-IN-EOR	34.6	1	11.0	1
C060	1·1-IN-ENOR	34.6	1	11.0	1
C061	2·1-IN-AND-EOR	22.7	1	—	—
C062	2·1-IN-AND-EOR	22.7	1	—	—
C063	2·1-IN-AND-EOR	34.6	2	11.0	1

NEC Electronics

CUSTOM/SEMICUSTOM

C064	1 · 1 · 1-IN-EOR	79.9	3	25.7	3
C065	1 · 1 · 1-IN-ENOR	79.9	3	25.7	3
C066	2 · 1-IN-AND-EOR	34.6	2	11.0	1
C067	1 · 2 · 2 · 2 · 2-IN-EXOR	—	—	13.1	3
C116	1 · 2 · 2-IN-AND-DFF	105.0	5	27.8	5
C117	2 · 3 · 3 · 2-IN-AND-DFF	110.4	5	—	—
C118	1 · 2 · 2-IN-AND-DFF	96.7	4	—	—
C119	2 · 3 · 3 · 2-IN-AND-DFF	107.5	5	—	—
C137 E131	MASTER-SLAVE-DFF	34.6	3	11.0	3

**OUTPUT BUFFER**

Block Name	Function	6301		6310/6320	
		Typical Power (mW)	Cell Count	Typical Power (mW)	Cell Count
G017	1-IN-AND	31.5	1	31.5	1
GA00	1-IN-NAND	31.5	1	31.5	1
GA01	2-IN-AND	36.0	1	36.0	1
GA02	2-IN-NAND	36.0	1	36.0	1
GA03	3-IN-AND	40.5	1	—	—
GA04	3-IN-NAND	40.5	1	—	—
GA05	4-IN-AND	45.0	1	—	—
GA06	4-IN-NAND	45.0	1	—	—
GA07	1-IN-AND/NAND	31.5	2	31.5	2
GA08	2-IN-AND/NAND	36.0	2	36.0	2
G006	3-IN-AND/NAND	40.5	2	—	—

**OUTPUT E.F. CIRCUIT**

Block Name	Function	6301		6310/6320	
		Typical Power (mW)	Cell Count	Typical Power (mW)	Cell Count
B230	OUTPUT LEVEL TRANSRATOR	18.2	—	18.2	—



**Description**

The μPB6101/6102/6103 are high-speed, low-power TTL compatible gate arrays using an advanced NEC bipolar technology. They consist of an uncommitted configurable logic array and TTL compatible input/output (I/O) buffers.

NEC's comprehensive CAD support system and master slice system significantly reduce the time and expense normally associated with semi-custom devices.

Normal turnaround time, after logic validation, is only 8-10 weeks. Advanced CAD tools, such as: logic simulation, automatic placement and routing, delay simulation, and test program generation are used to insure accurate error-free designs of all NEC gate arrays.

**Features**

- Choice of three (3) Array Sizes
- High Speed 2.5nS/gate
- Low Power 1.4 mW/gate
- Comprehensive CAD for Rapid, Error-Free Designs
- Remote CAD Access on the GE Mark III Time Share Network Using Customer Terminal for Logic Simulation
- Four Output Buffer Types Available (Totem Pole, Open Collector, Three-State, Bidirectional)
- Package Options Dual-in-Line packages, Flat-Packages, Pin-Grid Arrays
- 8-10 Week Delivery After Logic Simulation

**Packaging**

The package type (dual-in-line, flat package, or pin-grid array) is generally determined by the customer, while the number of pins is a function (sum) of the utilized I/O and power pins. The μPB610X Series can be packaged in a 16-pin DIP (μPB6101) (minimum) through 72-pin (μPB6103) (maximum) PGA

The Series consists of three (3) devices:

Part Number	Gates	Output Buffers	Input Buffers
μPB6101	256	32	38
μPB6102	598	52	64
μPB6103	918	69	69

**Package Pin Counts**

Part Number	Package Pins	
	Minimum	Maximum
μPB6101	16	40
μPB102	20	72
μPB6103	20	72

**ABSOLUTE MAXIMUM RATINGS**

ITEM	SYMBOL	RATING	UNIT
Power Supply Voltage	V <sub>CC</sub>	+7.0	V
Input Voltage	V <sub>I</sub>	-0.5 to +7.0	V
Output Voltage	V <sub>O</sub>	-0.5 to +5.5	V

**RECOMMENDED OPERATING CONDITION**

ITEM	SYMBOL	CONDITION	UNIT
Power Supply	V <sub>CC</sub>	5 ± 10 %	V
Operating Temperature	T <sub>OP</sub>	0 to 85	°C
"H" Level Output Current	I <sub>OH</sub>	100 MAX.	μA
"L" Level Output Current	I <sub>OL</sub>	12 MAX.	mA

**ELECTRICAL CHARACTERISTICS**  
D.C. CHARACTERISTICS ( $T_a=25^\circ\text{C}$ )

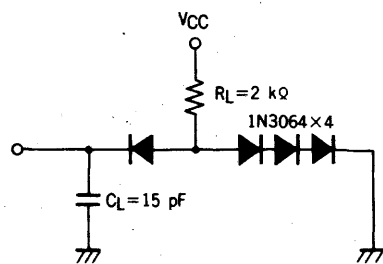
ITEM	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
Low Level Input Voltage	$V_{IL}$				0.8	V
High Level Input Voltage	$V_{IH}$		2.0			V
Input Clamp Voltage	$V_{IC}$	$V_{CC}=4.5\text{ V}, I_{IL}=-18\text{ mA}$	-1.5			V
Low Level Output Voltage	$V_{OL}$	$V_{CC}=4.5\text{ V}, I_{OL}=12\text{ mA}$		0.3	0.5	V
High Level Output Voltage	$V_{OH}$	$V_{CC}=4.5\text{ V}, I_{OH}=1\text{ mA}$ Note 1	2.5	3.0		V
Short Circuit Output Current	$I_{OS}$	$V_{CC}=5.5\text{ V}, V_O=0\text{ V}$	-100		-25	mA
Low Level Input Current	$I_{IL}$	$V_{CC}=5.5\text{ V}, V_I=0.4\text{ V}$	-200			$\mu\text{A}$
High Level Input Current	$I_{IH}$	$V_{CC}=5.5\text{ V}, V_I=2.7\text{ V}$			20	$\mu\text{A}$
High Level Output Current	$I_{OH}$	$V_{CC}=5\text{ V}, V_O=5.5\text{ V}$			100	$\mu\text{A}$
Off State Output Current	$I_{OZ}$	$V_{CC}=4.5\text{ V}, V_O=0.4\text{ V}/2.4\text{ V}$ (3 State Output)	-20		20	$\mu\text{A}$
Off State Output Current	$I_{OZ}$	(Bidirectional)	-200		40	$\mu\text{A}$

Note 1: Does not apply to Open-Collector Outputs

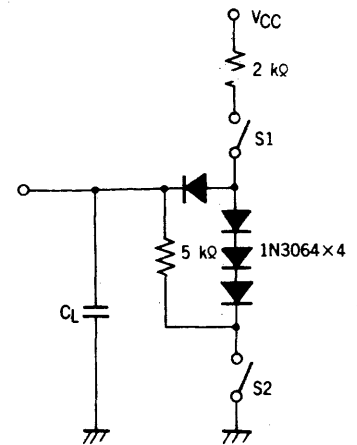
A.C. CHARACTERISTICS ( $T_a=25^\circ\text{C}$ )

ITEM	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
Delay Time/Gate	$t_{PHL}$	F/O=1	1.0	1.8	2.7	ns
	$t_{PLH}$	I=2 mm	1.4	2.5	3.7	ns
Delay Time/Input Buffer	$t_{PHL}$	F/O=1	1.0	1.8	2.7	ns
	$t_{PLH}$	I=2 mm	1.3	2.3	3.5	ns
Delay Time/Output Buffer	$t_{PHL}$	$C_L=15\text{ pF}$	2.8	4.7	7.1	ns
	$t_{PLH}$	$C_L=15\text{ pF}$	3.2	5.4	8.1	ns

OUTPUT LOAD CONDITION



i) for ordinary output



ii) for Three state output

EXAMPLE OF POWER DISSIPATION

ITEM	SYMBOL	POWER DISSIPATION	UNIT
Internal Gate	F010	1.4	mW
Input Buffer	M001	1.9	mW
Output Buffer (Totem-Pole)	B001	2.4	mW
Output Buffer (Open Collector)	B002	2.4	mW
Output Buffer (Three State)	B003	2.5	mW
Output Buffer (Bidirectional)	B004	4.4	mW
Output Buffer (Bidirectional D.C.)	B007	4.4	mW

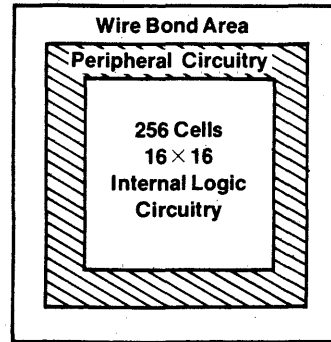
**Chip Architecture**

The arrays are divided into three basic areas. The I/O buffers are placed uniformly on the periphery of each device so that there are an equal number on each die edge. At each signal pad location, the pin function can be one of several possible I/O function configurations.

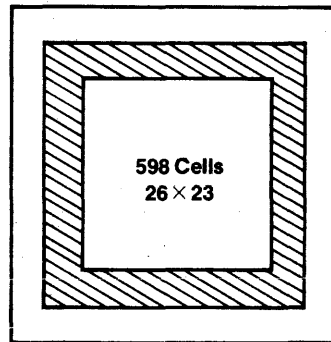
The area between the cells and the bonding pads contains all of the peripheral circuitry which convert from I/O levels to the internal logic, and a voltage generating for the internal power supply.

The center of the die contains rows of continuous gates separated by horizontal routing channels. The density of the array is a factor of the total cell count (row  $\times$  line). Each cell is made up of two transistors and four resistors.

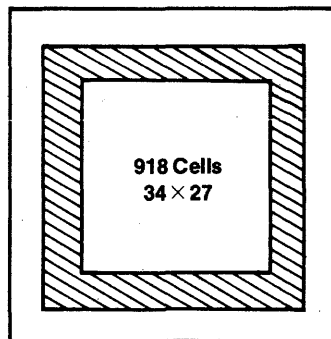
$\mu$ PB6101



$\mu$ PB6102



$\mu$ PB6103



NEC Electronics

CUSTOM/SEMICUSTOM

Fig. 1 Function block library list

BLOCK NAME	FUNCTION	CELL NO.	P <sub>D</sub> (mW)
F004	INVERTER	1	1.4
F000	2-INPUT NAND	1	1.4
F010	3-INPUT NAND	1	1.4
C020	6-INPUT NAND	2	1.4
C030	9-INPUT NAND	3	1.4
F011	3-INPUT AND	2	2.3
C021	6-INPUT AND	2	2.3
C51A	2-WIDE 3-INPUT AND OR INVERTER	2	2.8
C054	4-WIDE 3-INPUT AND OR INVERTER	4	5.6
F086	EXCLUSIVE OR	6	8.4
F074	D-F/F	6	8.4
C109	J-K F/F	7	9.8
C075	D-LATCH	4	5.6
CC10	3-STATE CONTROL	1	1.4
M001	INPUT BUFFER	-	1.9
B001	OUTPUT BUFFER (TOTEM-POLE)	-	2.4
B002	OUTPUT BUFFER (OPEN COLLECTOR)	-	2.4
B003	OUTPUT BUFFER (3-STATE)	-	2.5
B004	OUTPUT BUFFER (BIDIRECTIONAL)	-	4.4
B007	OUTPUT BUFFER (BIDIRECTIONAL) (OPEN COLLECTOR)	-	4.4
V000	0 LEVEL GENERATOR	1	1.4
JW1	WIRED AND	-	-
C005	DELAY GATE	4	3.7
CS10	OUTPUT LEVEL SHIFT	1	2.5

**DEVELOPMENT PROCEDURE**

Chart 1 illustrates the development step of both customer and NEC. The following two types of interfaces are available for customers.

1) CAD interface (Interface A) ..... most common method.

In this method the customer will begin at the logic design stage and proceed through to the logic simulation stage. NEC will then carry on from automatic placement and routing through sample delivery to final shipment of product.

The major benefit of this method is that the customer can perform the logic simulation on their own premises by using the NEC CAD facilities through the GE MARK III time share system, thus reducing the time needed for design.

Document

- (1) contact and non-disclosure agreement
- (2) circuit drawing based on NEC's functional block
- (3) interconnection data file (LOGINC)
- (4) test data pattern file (LOGPAT)
- (5) Pin-assignment (if required)
- (6) Critical path assignment (if required)

2) Circuit drawing interface (interface B)

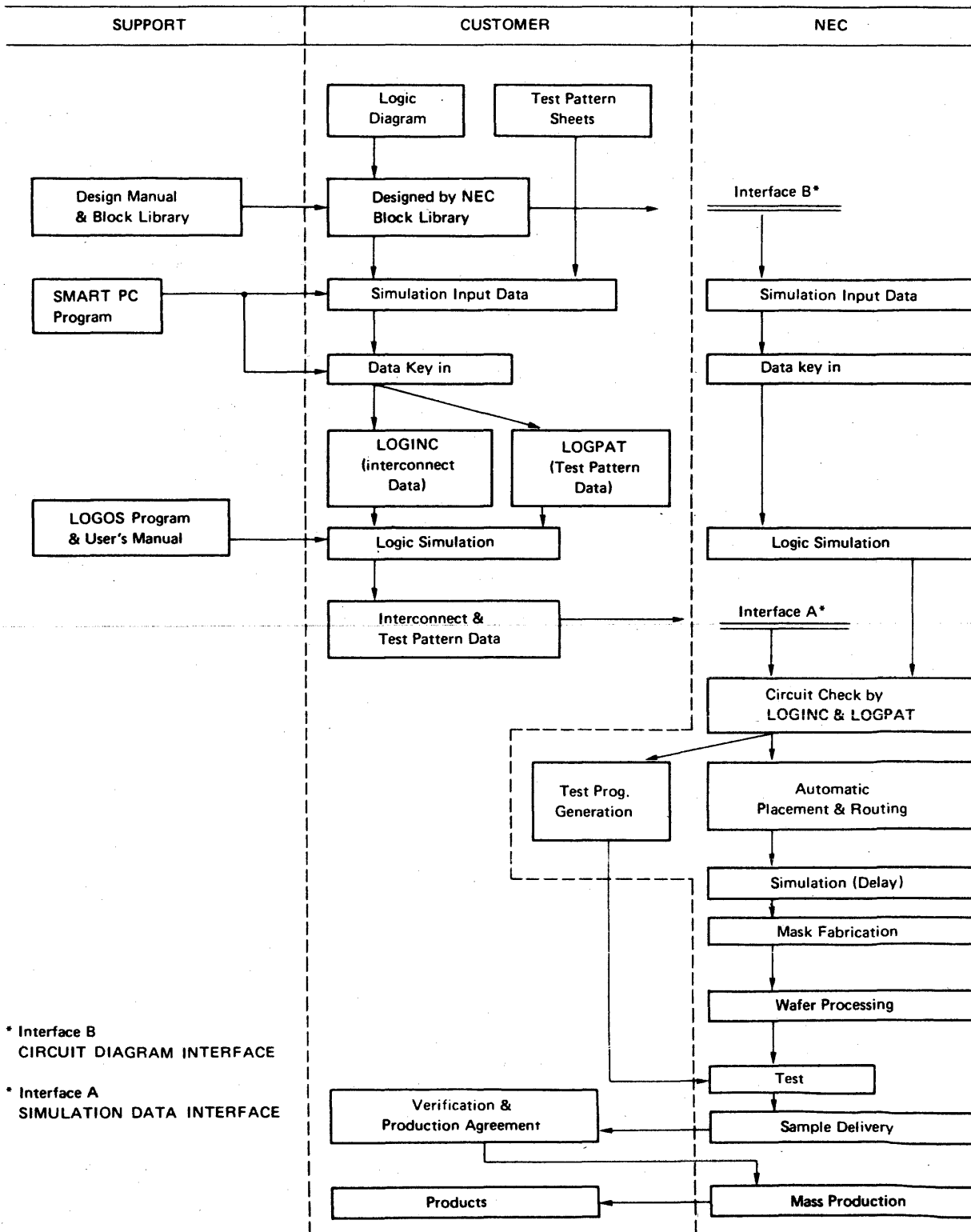
(1) logic drawing ..... Here the customer will prepare the logic drawing according to the block library shown in Fig. 1 and the design manual.  
(a typical example is shown in Fig. 2)

(2) test pattern sheet. .... The test pattern sheet also needs to be prepared  
(a typical example is shown in Fig. 3)

The remaining stages from logic simulation to shipment are performed by NEC. Design rule check and logic simulation results are reported to the customers.

NEC will furnish evaluation samples within 10 weeks of logic validation.

Flow Chart of Development and Interface



\* Interface B  
CIRCUIT DIAGRAM INTERFACE

\* Interface A  
SIMULATION DATA INTERFACE

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NEC Electronics

**Description**

The μPD 65000 Series is a family of low power, high-speed gate arrays manufactured using advanced Silicon gate CMOS technology. The series consists of from 427 to 11,250 gate devices that are available in a variety of package types.

The semi-custom approach of using Gate Arrays offers a unique and effective method of manufacturing I.C.'s with greatly reduced cost and development time. NEC makes this possible by stocking wafers which are completely fabricated except for the final step of interconnection of components to make up the required circuit configuration. This enables a designer the freedom of interconnecting the uncommitted components to achieve his unique circuit configuration.

The semi-custom approach is intended for customers seeking cost effective alternatives. By using Gate Arrays, customers can reduce component count and board size so that they can be more competitive in the markets which they serve. NEC's Gate Array program allows a customized I.C. to be developed with a short turn around time and at a small fraction of the cost of a full custom development program.

An advanced Computer-Aided Design (CAD) system is provided by NEC to achieve high quality, low cost and fast semi-custom I.C. development. Two kinds of interface are available for the customer. Interface "A" requires the customer to do the logic simulation either at one of the NEC's design centers or at his own facility through a terminal linked to G.E. Mark III time-sharing system. Interface "B" only requires the customer to provide the logic diagram and timing chart and NEC will do the rest.

So, contact us now and let NEC's Gate Array help you to succeed in your business.

Part Number	Number of Gates	Configuration	Input Buffers	Output Buffers
3-Micron Geometry				
μPD65002	858	66 × 13	50	50
μPD65003	427	61 × 7	38	36
μPD65010	1368	76 × 18	64	64
μPD65020	2112	96 × 22	80	80
2-Micron Geometry				
μPD65040	4104	152 × 27	120	116
μPD65060	6528	192 × 34	148	138
μPD65100	11,250	250 × 45	196	176

Notes:

1. A gate consists of 2 P-ch and 2 N-ch transistors.

**Features**

- |  | 3-micron  | 2-micron  |
|--|-----------|-----------|
| <input type="checkbox"/> High Speed  | 3ns/gate  | 2ns/gate  |
| <input type="checkbox"/> Low Power   | 30μW/gate | 20μW/gate |
| <input type="checkbox"/> Comprehensive CAD for rapid, error-free designs   |           |           |
| <input type="checkbox"/> Remote CAD access on the GE Mark III time share network using customer terminal for logic simulation. |           |           |
| <input type="checkbox"/> Four output buffer types available (normal, open drain, three state, bidirectional).                  |           |           |
| <input type="checkbox"/> Quick turnaround time (10~12 weeks)   |           |           |
| <input type="checkbox"/> Fully supported by advanced CAD   |           |           |
| <input type="checkbox"/> 20K gate automatic placement and routing capability   |           |           |
| <input type="checkbox"/> Logic simulation  |           |           |
| <input type="checkbox"/> Test program generation   |           |           |
| <input type="checkbox"/> Delay simulation  |           |           |

**ABSOLUTE MAXIMUM RATINGS**

(T<sub>A</sub> = 25°C)

ITEM	SYMBOL	RATING	UNIT
Power Supply Voltage	V <sub>DD</sub>	-0.5 ~ 7	V
Input Voltage	V <sub>I</sub>	-0.5 ~ V <sub>DD</sub> + 0.5	V
Output Current	I <sub>O</sub>	10	mA
Operating Temperature	T <sub>OPT</sub>	-40 ~ +85	°C
Storage Temperature	T <sub>STG</sub>	-65 ~ +150	°C

**Chip Structure**

The LSI chip consists of two areas; an internal cell area and a peripheral area.

**Input/Output Buffer**

All I/O buffers can be configured as input, schmit trigger, input, normal output, open drain output, three state output or bidirectional.

All I/O buffers are fully CMOS/TTL compatible.

All assigned outputs are capable of driving 8 LS-TTL loads. Additional driving capability can be obtained by connecting output buffers in parallel.

**Internal Cell**

One cell has 2 p-channel transistors and 2 n-channel transistors, which are arranged in an array surrounded by routing channels

One cell can be used as 2 input NAND gate so the number of cells is equal to the number of gates.

**Function Block**

LSI logic function can be realized using the internal cell area. SSI/MSI logic functions are configured in units of circuit assemblage called "block" or "macro" which can be placed in any location of the internal cell area. NEC has over 140 kinds of macro's. This enables flexible LSI design and reduces design TAT.

**Package Availability**

Plastic and ceramic DIP packages range from 40 pins to 64 pins, flat packages from 64 pins to 100 pins, and Pin Grid Arrays (PGA) packages from 72 pins to 208 pins.

A customer can select a specific package type according to design and I/O pin requirement.

**RECOMMENDED OPERATING CONDITION**

( $T_A = 25^\circ C$ )

ITEM	SYMBOL	TEST CONDITION	MIN	TYP	MAX	UNIT
Power Supply Voltage	$V_{DD}$	CMOS Level	4.5	5	5.5	V
Input Voltage	$V_I$		0	$V_{DD}$		V
Low Level Input Voltage	$V_{IL}$	CMOS Level	0	0.3 $V_{DD}$		V
High Level Input Voltage	$V_{IH}$	CMOS Level	0.7 $V_{DD}$	$V_{DD}$		V
Low Level Input Voltage	$V_{IL}$	TTL Level*	0	0.8		V
High Level Input Voltage	$V_{IH}$	TTL Level*	2.0	$V_{DD}$		V
Input Rise, Fall Time	$T_{RT}$		0	10		μS

\* $T_A = 0 \sim 70^\circ C$ ,  $V_{DD} = 5 V \pm 5\%$

**ELECTRICAL CHARACTERISTICS**

( $V_{DD} = 5 V \pm 10\%$ ,  $T_A = -40 \sim +85^\circ C$ )

ITEM	SYMBOL	CONDITION	MIN	TYP	MAX	UNIT
Static Current	$I_L$	$V_I = V_{DD}$ OR GND	0.1	200		μA
Dynamic Current	$I_{DD}$	1 MHz/cell	4			μA
Input Current	$I_L$	$V_I = V_{DD}$ OR GND	$10^{-5}$	10		μA
Low Level Output Current	$I_{OL}$	$V_{OL} = 0.4 V$	3.2	6		mA
High Level Output Current	$I_{OH}$	$V_{OH} = V_{DD} - 0.4 V$	1	2		mA
Low Level Output Voltage	$V_{OL}$	$I_O = 0$		0.1		V
High Level Output Voltage	$V_{OH}$	$I_O = 0$	$V_{DD}$		-0.1	V

**AC CHARACTERISTICS**

( $V_{DD} = 5 V \pm 10\%$ ,  $T_A = -40 \sim 85^\circ C$ )

ITEM	SYMBOL	CONDITION	MIN	TYP	MAX	UNIT
Max Operating Frequency	$F_{MAX}$		20	50		MHz
Delay Time *	$T_{PD}$	GATE	2			ns
	$T_{PD}$	I/O Buffer	12**			ns
Output Rise Time	$T_R$	$C_L = 15 pF$	10			ns
Output Fall Time	$T_F$	$C_L = 15 pF$	7			ns

\*With fan out of 3 and 3 mm wiring.

\*\*With 15pF

NEC Electronics

CUSTOM/SEMICUSTOM

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| <input type="checkbox"/> 20K gate automatic placement and routing capability   |           |           |
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| <input type="checkbox"/> Delay simulation  |           |           |

**ABSOLUTE MAXIMUM RATINGS**

(T<sub>A</sub> = 25°C)

ITEM	SYMBOL	RATING	UNIT
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Input Voltage	V <sub>I</sub>	-0.5 ~ V <sub>DD</sub> + 0.5	V
Output Current	I <sub>O</sub>	10	mA
Operating Temperature	T <sub>OPT</sub>	-40 ~ +85	°C
Storage Temperature	T <sub>STG</sub>	-65 ~ +150	°C



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A customer can select a specific package type according to design and I/O pin requirement.

**RECOMMENDED OPERATING CONDITION**

( $T_A = 25^\circ C$ )

ITEM	SYMBOL	TEST CONDITION	MIN	TYP	MAX	UNIT
Power Supply Voltage	$V_{DD}$	CMOS Level	4.5	5	5.5	V
Input Voltage	$V_i$		0	$V_{DD}$		V
Low Level Input Voltage	$V_{iL}$	CMOS Level	0	0.3	$V_{DD}$	V
High Level Input Voltage	$V_{iH}$	CMOS Level	0.7	$V_{DD}$		V
Low Level Input Voltage	$V_{iL}$	TTL Level*	0	0.8		V
High Level Input Voltage	$V_{iH}$	TTL Level*	2.0	$V_{DD}$		V
Input Rise, Fall Time	$T_{RT}$		0	10		μS

\* $T_A = 0 \sim 70^\circ C$ ,  $V_{DD} = 5 V \pm 5\%$

**ELECTRICAL CHARACTERISTICS**

( $V_{DD} = 5 V \pm 10\%$ ,  $T_A = -40 \sim +85^\circ C$ )

ITEM	SYMBOL	CONDITION	MIN	TYP	MAX	UNIT
Static Current	$I_L$	$V_i = V_{DD}$ OR GND	0.1	200		μA
Dynamic Current	$I_{DD}$	1 MHz/cell	4			μA
Input Current	$I_L$	$V_i = V_{DD}$ OR GND	$10^{-5}$	10		μA
Low Level Output Current	$I_{OL}$	$V_{OL} = 0.4 V$	3.2	6		mA
High Level Output Current	$I_{OH}$	$V_{OH} = V_{DD} - 0.4 V$	1	2		mA
Low Level Output Voltage	$V_{OL}$	$I_o = 0$			0.1	V
High Level Output Voltage	$V_{OH}$	$I_o = 0$		$V_{DD}$	-0.1	V

**AC CHARACTERISTICS**

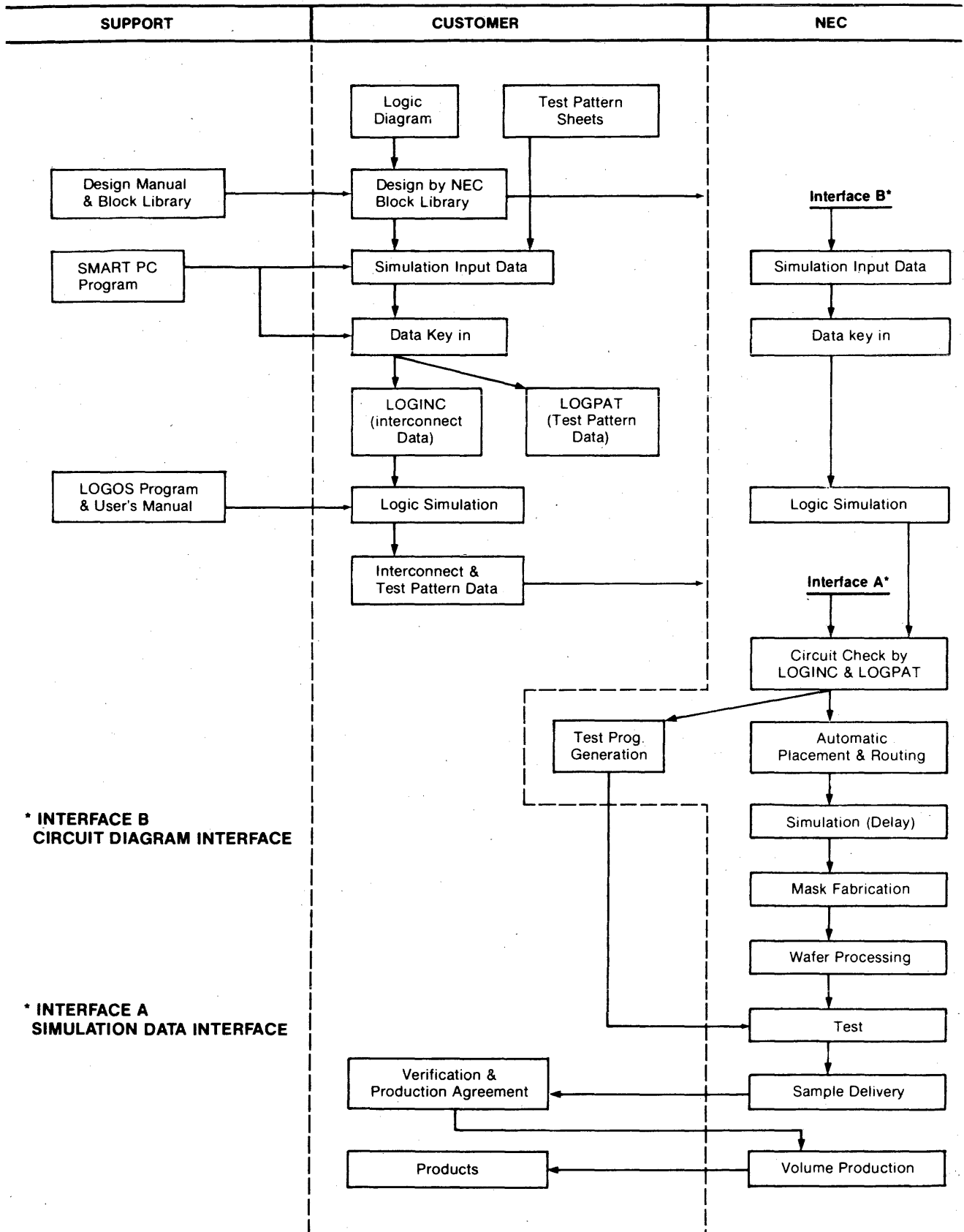
( $V_{DD} = 5 V \pm 10\%$ ,  $T_A = -40 \sim 85^\circ C$ )

ITEM	SYMBOL	CONDITION	MIN	TYP	MAX	UNIT
Max Operating Frequency	$F_{MAX}$		20	50		MHz
Delay Time*	$T_{PD}$	GATE	2			ns
	$T_{PD}$	I/O Buffer	12**			ns
Output Rise Time	$T_R$	$C_L = 15 pF$	10			ns
Output Fall Time	$T_F$	$C_L = 15 pF$	7			ns

\*With fan out of 3 and 3 mm wiring.

\*\*With 15pF

Chart 1. Flowchart of Development and Interface



\* INTERFACE B  
CIRCUIT DIAGRAM INTERFACE

\* INTERFACE A  
SIMULATION DATA INTERFACE

Figure 1. μD65002, μPD65010, μPD65020 Block List

INTERFACE BLOCK		
Block Type	Function	
F101	Input Buffer (CMOS Level)	
F102	Input Buffer (TTL Level)	
FO01	Output Buffer (Normal)	
EXT1	Output Buffer (Nch Open Drain)	
EXT2	Output Buffer (Pch Open Drain)	
B003	I/O Buffer (3 state CMOS Level)	
B004	I/O Buffer (3 state TTL Level)	
FUNCTION BLOCK		
Block Type	Function	Cells
IN-VERTER	F101 1-Input (F.O. = 5)	1
	F102 1-Input (F.O. = 10)	1
	F103 1-Input (F.O. = 15)	2
	F104 1-Input (F.O. = 20)	2
BUFFER	F111 1-Input (F.O. = 5)	1
	F112 1-Input (F.O. = 10)	2
	F113 1-Input (F.O. = 15)	2
	F114 1-Input (F.O. = 20)	3
NOR	F202 2-Input NOR Gate	1
	F203 3-Input NOR Gate	2
	F204 4-Input NOR Gate	2
OR	F212 2-Input OR Gate	2
	F213 3-Input OR Gate	2
	F214 4-Input OR Gate	3
NAND	F302 2-Input NAND Gate	1
	F303 3-Input NAND Gate	2
	F304 4-Input NAND Gate	2
	F305 5-Input NAND Gate	3
	F306 6-Input NAND Gate	3
AND	F312 2-Input AND Gate	2
	F313 3-Input AND Gate	2
	F314 4-Input AND Gate	3
AND-NOR	F421 2 + 1 Input AND-NOR Gate	2
	F422 2 + 2 Input AND-NOR Gate	2
	F423 3 + 1 Input AND-NOR Gate	2
	F424 2 AND, 2-NOR	2
OR-NAND	F431 2 + 1 OR-NAND Gate	2
	F432 2 + 2 OR-NAND Gate	2

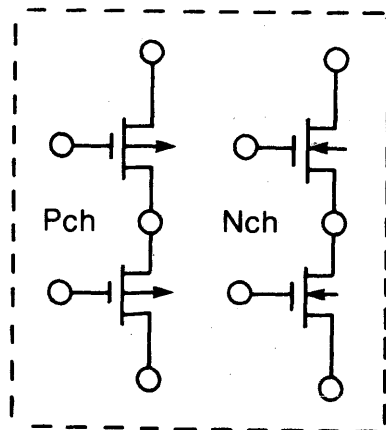
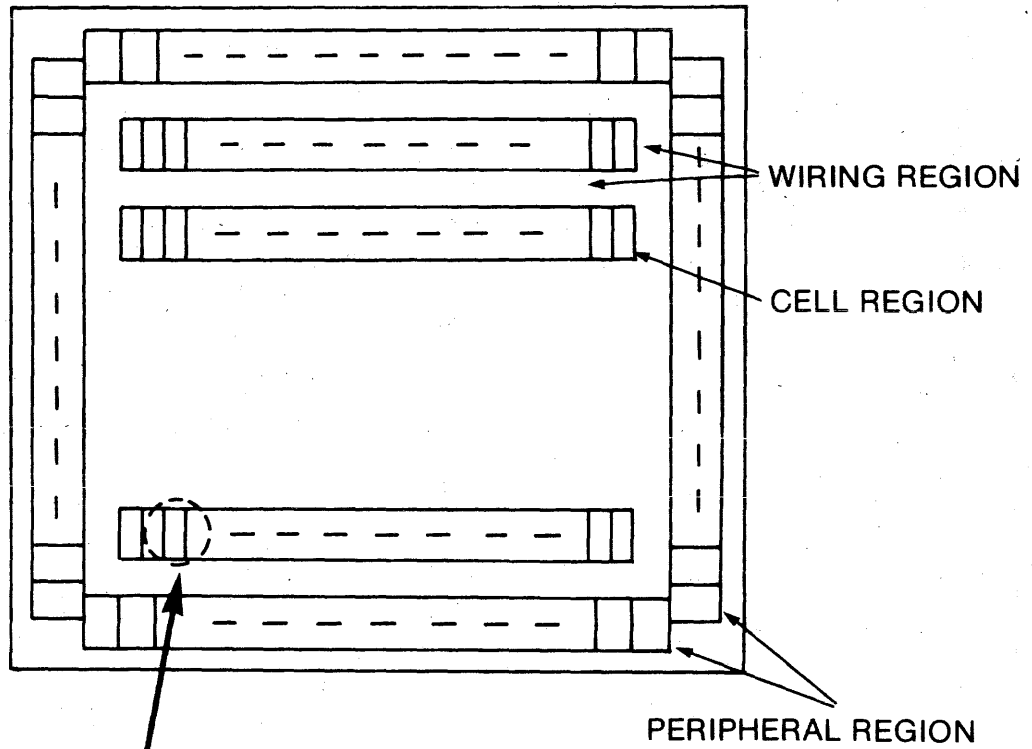
FUNCTION BLOCK		
Block Type	Function	Cells
	F433 3 + 1 OR-NAND Gate	2
	F434 2 OR, 2 NAND	2
DRIVER	F501 Clock Driver	1
	F502 Clock Driver (Dual)	2
	F503 Clock Driver (Parallel)	3
	F504 Clock Driver (Pala-Dual)	4
	F505 OSC Control	2
	F531* 3-State Buffer (EN)	3
	F532* 3-State Buffer (EN)	3
EX-OR	F511 2-Input Exclusive OR Gate	3
EX-NOR	F512 2-Input Exclusive NOR Gate	3
LATCH	F601 D-Latch	3
	F602 D-Latch with Reset	4
	F603 D-Latch with Reset	4
	F604* D-Latch C	3
	F605* D-Latch C with Reset	4
FLIP-FLOP	F611 D-Type	5
	F612 D-Type with Reset	7
	F613 D-Type with Set	7
	F614 D-Type with Set-Reset	7
	F615* D-Type with Reset	7
	F616* D-Type with Set	7
	F617* D-Type with Set-Reset	7
	F631* D-Type C	5
	F635* D-Type C with Reset	7
	F636* D-Type C with Set	7
	F637* D-Type C with Set-Reset	7
	F641* D-Type (Buffered Out)	6
	F642* D-Type (Buffered Out) with Reset	6
	F643* D-Type (Buffered Out) with Set	8
	F644* D-Type (Buffered Out) with Set-Reset	8
	F645* D-Type (Buffered Out) with Reset	8
	F646* D-Type (Buffered Out) with Set	8
	F647* D-Type (Buffered Out) with Set-Reset	8
	F661* D-Type (Buffered Out) with C	8

(\* Under Designing)

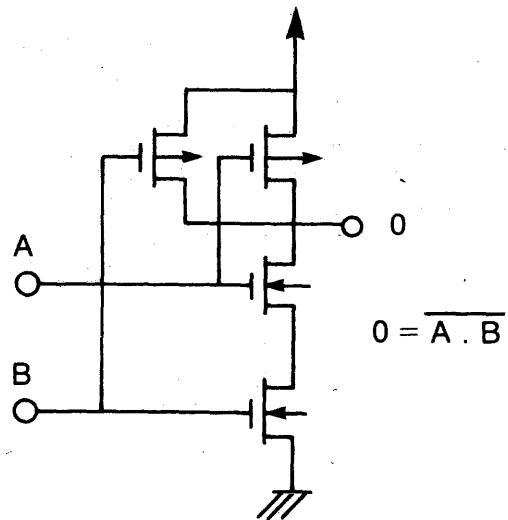
FUNCTION BLOCK		
Block Type	Function	Cells
	F665* D-Type (Buffered Out) with Reset	8
	F666* D-Type (Buffered Out) with Set	8
	F667* D-Type (Buffered Out) with Set-Reset	8
	F691 Serial/Parallel Shift Register	5
	F712 Toggle with Reset	7
	F713 Toggle with Set	7
	F714 Toggle with Set-Reset	7
	F715* Toggle with Reset	7
	F716* Toggle with Set	7
	F717* Toggle with Set-Reset	7
FLIP-FLOP	F735* Toggle (T) with Reset	7
	F736* Toggle (T) with Set	7
	F737* Toggle (T) with Set-Reset	7
	F742* Toggle (Buffered Out) with Reset	8
	F743* Toggle (Buffered Out) with Set	8
	F744* Toggle (Buffered Out) with Set-Reset	8
	F745* Toggle (Buffered Out) with Reset	8
	F746* Toggle (Buffered Out) with Set	8
	F747* Toggle (Buffered Out) with Set-Reset	8
	F766* Toggle (Buffered Out) (T) with Set	8
	F767* Toggle (Buffered Out) (T) w/Set-Reset	8
	F771* J-K F/F	9
	F772* J-K F/F with Reset	11
	F773* J-K F/F with Set	11
	F774* J-K F/F with Set-Reset	11
	F775* J-K F/F with Reset	11
	F776* J-K F/F with Set	11
	F777* J-K F/F with Set-Reset	11
	F781* J-K F/F C	9
	F785* J-K F/F C with Reset	11
	F786* J-K F/F C with Set	11
	F787* J-K F/F C with Set-Reset	11

(\* Under Designing)

**Chip Layout of CMOS Gate Array**



**CELL CONFIGURATION  
(1 CELL)**



**EXAMPLE OF  
GATE CONFIGURATION  
(2-IN NAND GATE)**



# **Don't Miss the IC Updates**

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Circuits Magazine**

# SCD4000 600 Gate ECL Array

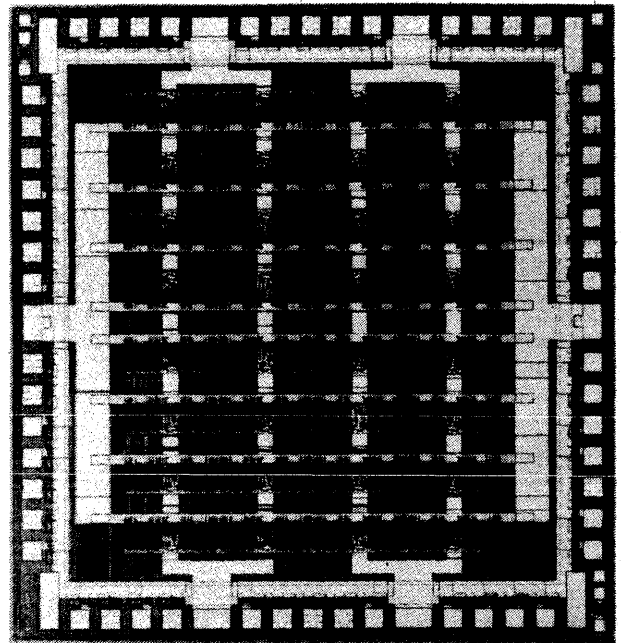
## General Description

The Plessey SCD4000 is part of a family of semi-custom gate arrays using emitter coupled logic (ECL) circuits fabricated on a high performance bipolar process to give very fast switching speeds. ECL technology offers the fastest possible gate arrays. SCD4000 achieves basic gate delays of 400ps and D-type flip-flop clock rates of 400MHz typical. The array can be used as an economical replacement for several ECL 10K and 100K standard SSI and MSI parts even at low volumes. The design procedure consists of customising two metallisation layers which are then applied to pre-processed slices held in stock to manufacture quickly a new circuit or follow-on order from the customer.

## Features

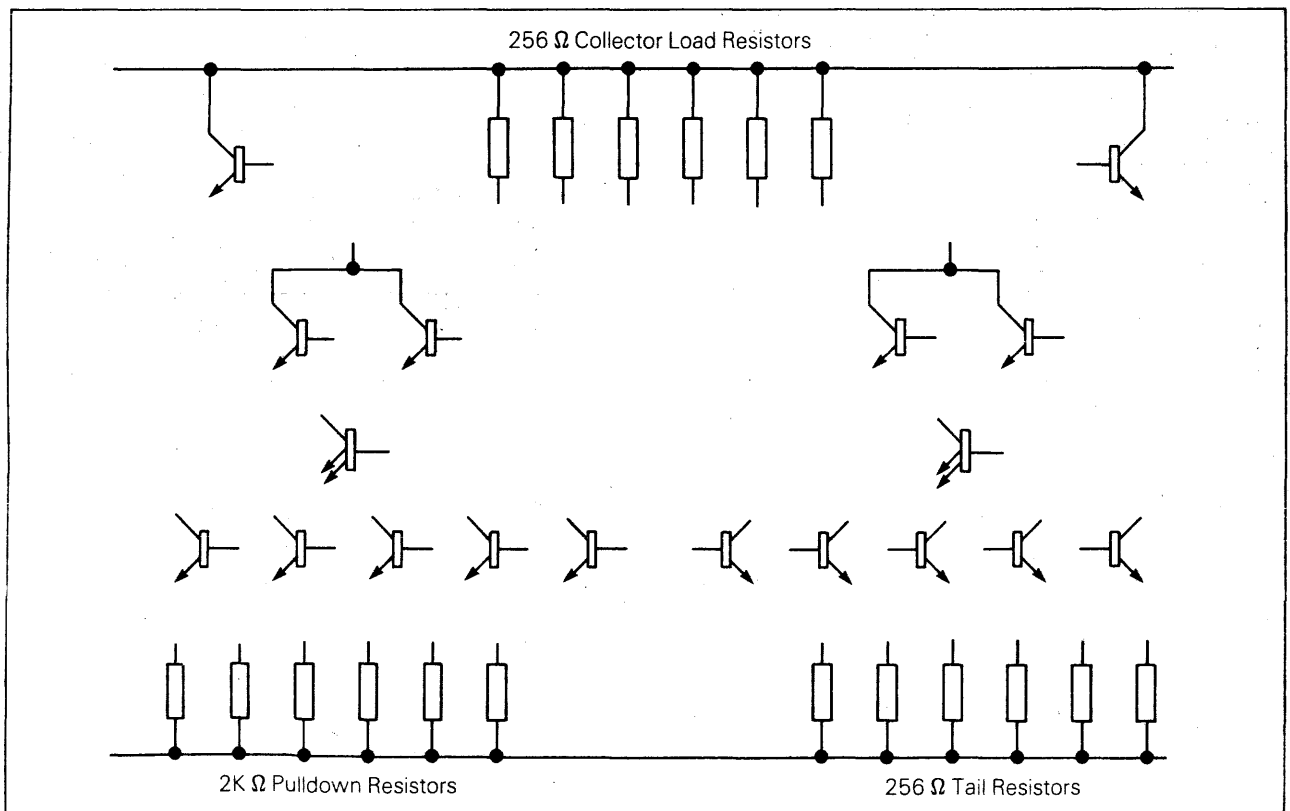
- Approx. 600 logic gate complexity.
- Compound (OR-AND-OR) gate delays typically 500ps.
- Basic gate delays typically 400ps.
- Series gated (or stacked) ECL logic uses reduced internal voltage swings (500mV).
- ECL 10K compatibility.
- Fully automated design procedure.
- Extensive library of standard cells.
- Two layer metallisation allows up to 100 per cent utilisation.
- 56 input/output pins, 10 power connection pins.
- Maximum power dissipation typically 3.5 watts (100 per cent utilisation).

Figure 1 SCD4000 Chip Photograph



(chip size 170 x 170 mil)

Figure 2 SCD4000 Cell Components



## Array Organisation

The array is configured into 8 x 8 cells each of which contains 20 transistors and 18 resistors as shown in figure 2. Each pair of cells shares a common voltage bias which provides all the reference levels required for internal ECL circuits and also for interfacing ECL 10K logic levels. Three of the reference levels are used for series gating. This enables complex functions to be implemented with a single cell.

The cell components have no predefined interconnect pattern. Consequently logic functions can be implemented using any required component configuration. Furthermore, additional logic is available by using the wired-AND and wired-OR functions. The result of using these powerful techniques is a 64 cell array which provides a conservative 600 logic gate complexity. (In fact, were the array to be implemented with high efficiency functions such as parity gates, a gate count of 960 would be achieved even before using the wired function).

The array has 64 edge pads of which 10 are reserved for power connections. The remaining 54 edge pads are freely available for either input or output functions as each pad site contains high current output transistors and input interface components.

## Power Dissipation

Clearly power dissipation is dependent both upon the complexity and implementation of a given design. A completely full array will have a power dissipation of about 3.5 watts. The use of series gated logic greatly reduces the power necessary to perform a logical function as complex functions (e.g. 4 to 1 multiplexer) can be implemented on a single current source.

## Cell Library Coding

The contents of the cell library are shown in Table 1. Each cell may contain more than one logic element and each element may have a number of options. For each cell there is a functional description showing logic levels, truth table and other circuit elements such as level shifters. A particular code is required to define the function of a specific element used within a cell. The format is as follows.

GE WW X Y Z

Where

- GE is the library cell.
- WW defines the cell.
- X selects gate or circuit element.
- Y defines variant option.
- Z input reference level.

## SCD4000 Design Procedure

SCD4000 is supported by a fully automated software package including CLASSIC (an event driven logic simulator) and SCARP (an autoroute program) together with comprehensive electrical and rule violation checks. The flow diagram in figure 3 indicates the typical design procedure.

The design route outlined here will enable customers to successfully produce an ECL gate array in typically 10-16 weeks. The customer does not need to acquire specialised integrated circuit knowledge to be able to design devices. A number of customer/Plessey interfaces are possible during the course of a given design, ranging from simple logic specification at one extreme to magnetic tape (containing all the information necessary to drive the maskmaking pattern generator) at the other.

Figure 3 Typical Design Flow Chart

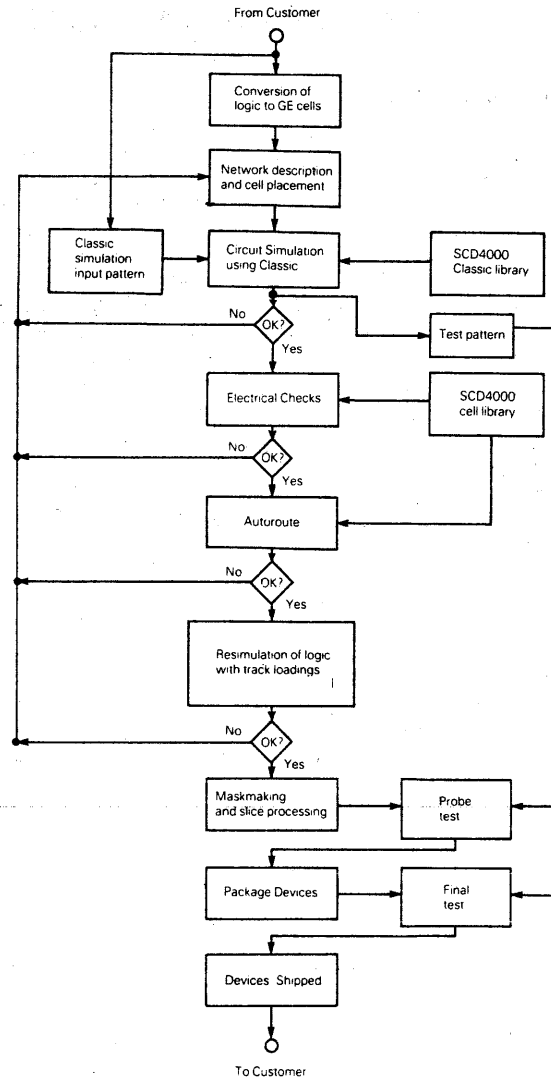


Table 1 GE Cell Library

Cell Name	Logic Function	Input/Output Buffer Compatible	
		Input	Output
GEAA	3 input buffer gate + 2 input OR	Yes	Yes
GEAB	Triple 2 input OR/NOR	Yes	Yes
GEAC	Dual 5 input OR/NOR	Yes	Yes
GEAN	Hex 1 input buffer (true and inverse)	Yes	Yes
GEAP	Quad 1 input buffer (true and inverse)	Yes	Yes
GEAT	Triple 3 input OR/NOR	Yes	Yes
GEBC	Dual comparator	No	No
GECA	Dual 2 input OR/3 input OR-AND	Yes	No
GECC	Dual 2 input OR/2 input OR-AND	No	Yes
GECC	4 input OR/3 input/3 input OR-AND	Yes	No
GEDF	2 to 4 decoder with force high	Yes	No
GEEB	Dual Equivalence/Not Equivalence	No	Yes
GEFA	Full Adder	No	No
GEMA	4 to 1 multiplexer inverter	No	No
GEMB	Dual multiplexer (true and inverse)	No	Yes
GEMF	Dual multiplexer with force high or low	No	No
GEP A	4 input parity	No	No
GESA	Dual D-latch + inverter	No	No
GESB	Dual D-latch (true and inverse)	No	No
GETF	3 to 1 multiplexer with force high or low	No	No
GETV	2 to 1 multiplexer master-slave latch	No	No

For further information please contact: *The I.C. Marketing Manager*

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## 2.5 MICRON CMOS GATE ARRAYS

This range of CMOS Gate Arrays adds an increased maximum size and speed capability to Plesseys existing Five and Four micron geometry products. They use the same well proven proprietary Plessey CAD system ensuring trouble free designs.

	CLA5100	CLA5200	CLA5300	CLA5400	CLA5500	CLA5600	CLA5700	CLA5800	CLA5900
Technology	Silicon Gate	Silicon Gate	Silicon Gate	Silicon Gate	Silicon Gate	Silicon Gate	Silicon Gate	Silicon Gate	Silicon Gate
Process Geometry ( $\mu\text{m}$ )	2.5	2.5	2.5	2.5	2.5	2.5	2.5	2.5	2.5
User-defined metal interconnection levels	2	2	2	2	2	2	2	2	2
Bonding Pads									
Uncommitted	36	48	64	80	96	112	128	144	160
Power	4	8	8	8	16	16	16	16	16
-Total	40	56	72	88	112	128	144	160	176
No. of equivalent 2 input NAND gates	640	1232	2016	3060	4404	5984	7104	8856	10,440
Typical gate utilization %	87	87	87	87	87	87	87	87	87
Gate delay (typ) (nS)									
Inverter F.0.1	0.9	0.9	0.9	0.9	0.9	0.9	0.9	0.9	0.9
3 I/P NAND F.0.3	2.8	2.8	2.8	2.8	2.8	2.8	2.8	2.8	2.8
System clock (MHz)	25	25	25	25	25	25	25	25	25
Toggle Freq.	50	50	50	50	50	50	50	50	50
Power consumption per gate (average)	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz	1 $\mu\text{W}$ /MHz
Power supply range (V)	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7
Ambient operating Temperature ( $^{\circ}\text{C}$ )	-55 to +125	-55 to +125	-55 to +125	-55 to +125	-55 to +125	-55 to +125	-55 to +125	-55 to +125	-55 to +125
I/O Compatibility A=CMOS B=TTL, C=LSTTL	ABC	ABC	ABC	ABC	ABC	ABC	ABC	ABC	ABC
Die size (mils)	111x111	138x138	166x166	194x194	235x235	269x268	291x291	319x324	346x346
Automatic Placement	YES	YES	YES	YES	YES	YES	YES	YES	YES
Automatic Routing	YES	YES	YES	YES	YES	YES	YES	YES	YES
Masks needed for customization	3	3	3	3	3	3	3	3	3

### AVAILABILITY

Plessey is now accepting designs for prototyping with full production capacity available second quarter 1984.





# PLESSEY Solid State

A Division of Plessey Trading Corporation  
3 Whatney, Irvine, California 92714  
Telephone (714) 951-5212 TWX 910-595-1930  
Telex: 701464 ANSBCK PLESSEY

## Semi-Custom Design

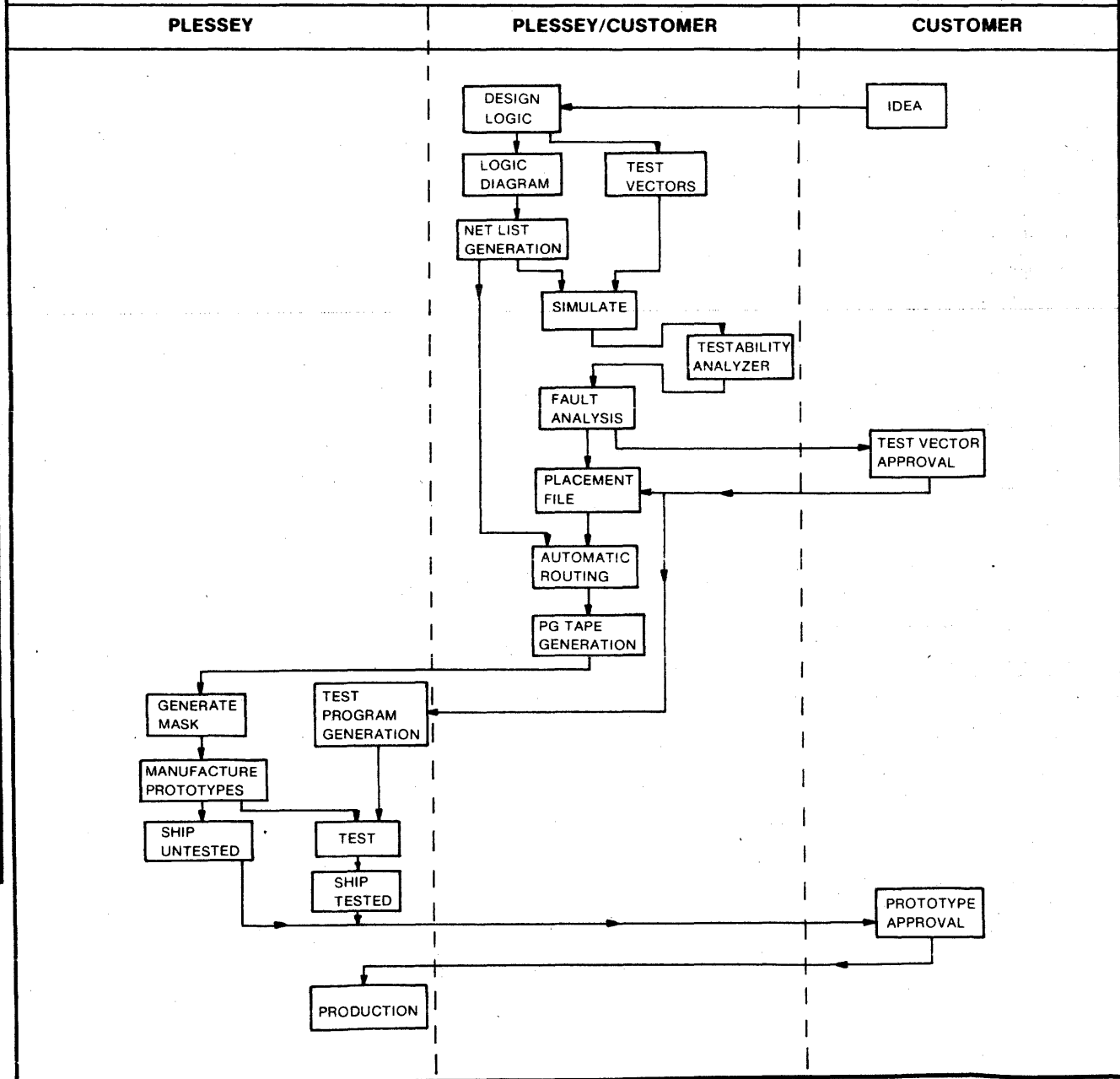
Plessey offers high density cell-based and Gate Array *Semi-Custom* devices fabricated on the best Bipolar, CMOS and NMOS technologies currently available. A full custom design and custom processing facility is also available.

The selection of the best approach for a par-

ticular problem depends upon a number of factors such as circuit size, development time and cost, and unit production cost amongst others.

Plessey will be pleased to assist in the selection of an optimal solution on receipt of complete information.

### PLESSEY SEMICONDUCTORS CMOS GATE ARRAYS DESIGN RESPONSIBILITIES



Plessey Semiconductors

CUSTOM/SEMICUSTOM

This series of CMOS Gate Arrays offer a quick and convenient method of producing low cost and high density customized integrated circuits.

This family of devices features very low power consumption using the latest silicon gate CMOS technology.

Two customizing layers of metallization give maximum design flexibility, speed and efficient use of chip area.

A sophisticated CAD system facilitates design and makes fully functional parts first time a certainty.

	CLA1000	CLA1200	CLA1500	CLA2100	CLA2300	CLA2500	CLA3700
TECHNOLOGY	SILICON GATE	SILICON GATE	SILICON GATE	SILICON GATE	SILICON GATE	SILICON GATE	SILICON GATE
PROCESS GEOMETRY ( $\mu\text{m}$ )	5	5	5	5	5	5	4
User-defined Metal Interconnection Levels	1 <sup>①</sup>	1 <sup>①</sup>	1 <sup>①</sup>	2	2	2	2
Bonding Pads - Uncommitted	38	50	62	40	52	64	80
- Power	2	2	2	4	4	4	4
- Total	40	52	64	44	56	68	84
No. of Equivalent 2-input NAND gates	560	960	1440	840	1440	2400	4200
Typical Gate Utilization	70	70	70	87	87	87	87
Gate Delays - Typical (ns)	6	6	6	4	4	4	3
Maximum	9	9	9	6	6	6	5
Toggle Frequency (MHz)	20	20	20	30	30	30	50
Power Dissipation @ gate	12 $\mu\text{W}$ /MHz	12 $\mu\text{W}$ /MHz	12 $\mu\text{W}$ /MHz	10 $\mu\text{W}$ /MHz	10 $\mu\text{W}$ /MHz	10 $\mu\text{W}$ /MHz	8 $\mu\text{W}$ /MHz
Typical Array Power Dissipation (mW)	30.6	52.4	78.6	47.5	81.5	135.7	190.0
Power Supply Range (V)	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7	3 to 7
Ambient Operating Temperature ( $^{\circ}\text{C}$ )	-55 to 125	-55 to 125	-55 to 125	-55 to 125	-55 to 125	-55 to 125	-55 to 125
I/O Compatibility A=CMOS,B=TTL,C=LSTTL	A,B,C	A,B,C	A,B,C	A,B,C	A,B,C	A,B,C	A,B,C
Die Size (mils)	191 x 131	211 x 177	223 x 233	196 x 132	243 x 156	243 x 230	233 x 244
Density (gates/mm <sup>2</sup> )	60	60	60	100	100	100	120
Automatic Placement	NO	NO	NO	YES <sup>②</sup>	YES <sup>②</sup>	YES <sup>②</sup>	YES <sup>②</sup>
Automatic Routing	NO	NO	NO	YES	YES	YES	YES
Masks Needed for Customization	1	1	1	3	3	3	3

- ① Plessey performs layout  
 ② Auto-placement to be implemented 3Q83



# CMOS MICROCELL

## MV2000 Series

### 2.5 MICRON DOUBLE LAYER METAL

#### Introduction

Microcell is a powerful blend of more conventional standard cell, distributed gates, and generated logic approaches to LSI circuit design. It uses Plessey's well established CMOS process and versatile double layer metal interconnect. Very high circuit speeds are possible with the 2 1/2 micron geometry used.

This approach allows circuit costs to approach those of full custom ICs without the time and cost penalties normally associated with full custom.

#### Design Process

The design process is arranged to allow the fullest possible checking to ensure first time success. The Microcell approach itself reduces the amount of checking to be done as all standard cells are proven and, hence, only the interconnections require checking in the simulation. Design is done using the same comprehensive and versatile CAD system developed over many years for Plessey's existing range of microcell and gate array products.

#### Specification 25°C

Gate delay (typ) (output falling)	
2 input NAND FO 1	0.8nS
2 input NAND FO 3	1.2nS
Maximum system clock	25MHz
Maximum toggle frequency	50MHz
Power consumption/gate (average)	1 W/MHz
Power supply	3 to 7v (5V preferred)
Operating temperature	-55 to +125°C
I/O Compatibility	CMOS, TTL, LSTTL
Store Temperature	-65 to +150°C

#### Packaging

Standard packages are as follows:

24, 28, 40, 48 pin DII 68, 84 pin LCC, 68, 84, 108, 132 PGA

Supply of devices in other packages is subject to negotiation. Below 24 pins standard DIL packages are standard.

Plessey Semiconductors  
CUSTOM/SEMICUSTOM

**Cell Library**

CTINV1	Small Inverter
CTINV4	Large Inverter
CTNAND2	Two Input Nand Gate
TNAND3	Three Input Nand Gate
CTNAND4	Four Input Nand Gate
CTNOR2	Two Input Nor Gate
CTNOR3	Three Input Nor Gate
CTNOR4	Four Input Nor Gate
CTAOI	And Or Inverter Cell
CTOAI	Or And Inverter Cell
CT2ANOR	Two Input And Nor Gate
CT2ONAND	Two Input Or Nand Gate
CT4ANOR	Four Input And Nor Gate
CT4ONAND	Four Input Or Nand Gate
CT42ANOR	Four and Two Input And Nor Cells
CT42ONAN	4 + 2 Input Or Nand Gate
CTEXOR	Two Input Exclusive Or
CTEXNOR	Two Input Exclusive Nor
CTHADD	Half Adder
CTFADD	Full Adder
CTCK1	Small Clock Generator
CTCK2	Medium Clock Generator
CTCK3	Large Clock Generator
CTDL	Latch
CTDLRS	Latch with Set and Reset
CTMSD	Master-slave D-type
CTMSDRS	Master-slave D-type with Set and Reset
CTSR	Shift Register Cell
CTSRRS	Shift Register Cell with Set and Reset
CTTRA	Line Driver
CT2TRA	Two into One Multiplexer
CTPAD	Pad cell used for supply connection etc.
CTIP	Input Pad
CTIPBN	Normal TTL Input Buffer
CTIPBL	Large TTL Input Buffer
CTPRO	I/P Protection Cell
CTRH+	High Value pull up +V
CTRH-	High Value pull down -V
CTRL+	Low Value pull up +V
CTRL-	Low Value pull down -V
CTSCHMIT	Schmitt Trigger
CTOPD	Output Pad
CTOPL	Large Output Driver
CTOPQ	Quad Output Driver
CTOPTRD	Output Pad with separate P and N devices
CTOPTRL	Large Tristate Driver
CTOPTRQ	Quad Tristate Output Driver
CTTRID	Tristate Driver
CTFILL1	Filler cell for spacing Inverter and similar height cells
CTFILL2	Filler cell for spacing CTEXOR and similar height cells
CTFILL3	Filler cell for spacing CTFADD and similar height cells
CTCKF	Clock Filler Cell
CTCKDS	Clock Filler Cell
CTCKFIL1	Clock Filler Cell
CTCKFIL2	Clock Filler Cell
CTSRFIL	Filler Cell for CTSR and CTSRRS

Other cells including PLA, ROM, RAM and a Microprocessor will be introduced in due course.

**Qualification**

Parts are normally supplied to Plessey's stringent commercial release standards. Screening to MIL883B and equivalent standards is available on request at additional cost.

# Eliminate Gate



## With Raytheon's Full Service Approach to Advanced Bipolar ISL Configurable Gate Array Problem Solving.

Raytheon

CUSTOM/SEMICUSTOM

Your reason for considering semicustom gate arrays is for the time and cost effective superior product solutions they provide. Raytheon achieves those goals for you, recognizing and avoiding the many potentially risk inducing pitfalls that could affect your objectives.

Employing service intensive customer applications support and proven automated design techniques, Raytheon strives for error-free gate array implementation the first time.

### Experience

Raytheon was among the very first to design and manufacture gate arrays. Our long standing dedication to hard-to-do, high-performance, high-reliability design and production requirements carries over to all aspects of our gate array programs.

We began with a 60-gate device in 1974, followed by a 300-gate device in 1978 and a bipolar ISL 1620-gate density array in 1981. Over 70 circuit designs have been successfully developed, and are now in volume production. We now have five members of our ISL CGA family, ranging from 800 to 2400-gates.

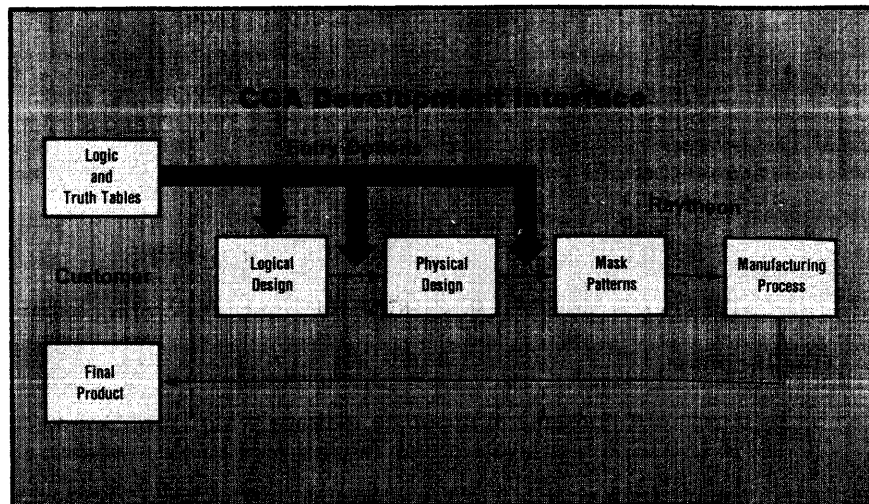
### Personalized Development Support

Raytheon's cost effective approach to semicustom circuit implementation makes it fast and easy to develop your prototypes. Working parts can be in your system within 8 weeks.

Raytheon can start with your logic diagram and truth tables or divide the design tasks with you to optimize cost, minimize the turnaround time, and ensure design control.

Regardless of the design entry point that you choose, the Raytheon design engineering team will work closely with your designers to help assure optimum product performance, cost-effective partitioning, accurate logic translation, and required testability.

A comprehensive design source manual and an intensive 3-day tutorial are available to those customers who will be performing the circuit design portions of development.



# Array Anxiety

## Design Automation

Raytheon's CAD facility offers a comprehensive assortment of analysis and design software, in a **common data base** environment, to quickly produce error-free prototype devices:

- Schematic Capture
- Logic Simulation
- Fault Grading
- Auto-Placement and Routing (Auto-Layout)
- Performance Verification

## Performance Characteristics

Raytheon ISL CGAs are all designed for radiation tolerance and to operate over the  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  temperature range. Full MIL-STD-883, Class B or S processing, program management, and other custom or special requirements can be satisfied.

## Bipolar ISL Technology

Raytheon's highly manufacturable 3 micron bipolar ISL process produces high density, superior performance, and efficient power consumption on all configurable gate arrays.

**Interdensity compatibility** exists between all arrays. Topological uniformity permits universal macrofunction and I/O designs.

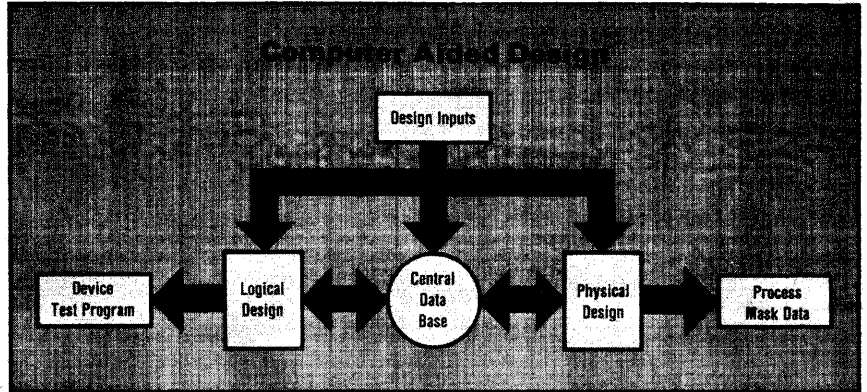
All Raytheon CGA's are fully TTL compatible. You can choose the desired function from a list of 35 I/O configurations and situate it at any signal pad location.

## Package Options

Everything will be done to put your array in the package you need. Choose from ceramic DIPs, leadless chip carriers, flatpaks, pin grid arrays, or name the custom package you require.

## Dedicated Service

Raytheon's high level of applications support begins before an order is received. Call us today. Let's examine your needs, and explore a thoroughly analyzed and properly executed configurable gate array development. Raytheon means service and dedication.



	Part Numbers				
	CGA8L48	CGA12L60	CGA16L68	CGA20L76	CGA24L84
Number of Array ISL Gates	836	1196	1620	1984	2376
Number of I/Os	48	60	68	76	84
Standard Speed (nS)	2.3	2.3	2.3	2.3	2.3
Enhanced Speed (nS)*	1.8	1.8	1.8	1.8	1.8
Power /Gate ( $\mu\text{W}$ )	350	350	350	350	350
Operating Power Supply ( $V_{CC}/V_{BB}$ )	5/1.8	5/1.8	5/1.8	5/1.8	5/1.8

\*A Raytheon proprietary feature that is available on all ISL gates.

The following MSI functions are a partial listing of those available in our **Macrofunction Library** for convenient implementation of customer circuit design.

Function	5400/7400 Series No.*
D-Type Flip-Flop with Clear and Preset	74
Two Input Exclusive-OR Gate	86
J-K Type Flip-Flop	109
3-to-8 Line Decoder	138
8-to-3 Line Priority Encoder	148
1-of-8 Data Selectors/Multiplexers with Strobed Outputs	151
4-to-1 Data Selector/Multiplexer	153
4-to-16 Line Decoder/Demultiplexer	154
Synchronous Decade Counter	162A
4-Bit Bidirectional Universal Shift Register	194
S-R Latch	279
9-Bit Odd/Even Parity Generator/Checker	280
4-Bit Binary Full Adder with Fast Carry	283
8-Bit Universal Shift/Storage Register	299
Octal Transparent Latches	373
Octal D-Type Edge-Triggered Flip-Flops	374
4-Bit ALU	381

\*The ISL logic format functions are pin polarity and function equivalents of the corresponding 5400/7400 LS Series functions. Raytheon reserves the right to change the circuitry and other data at any time without notice.



Raytheon Company, Semiconductor Division  
350 Ellis Street, Mountain View, CA 94043  
(415) 966-7716

# Custom and Semicustom LSI

## RCA Semicustom LSI Capabilities

### Automated GAs

#### Features

- Easy to Design and Change
- Fast Turnaround • CMOS Performance
- Wafer Stockpiling • CAD Compatible
- Library of Logic Macros

As gate counts and design complexities increase, manual layout of GAs becomes costly and time consuming. Therefore, RCA has developed automated placement and routing techniques.

The Automated GA System requires only one unique metal mask level. Artwork for this level is generated automatically from RCA's extensive library of logic cells.

The metal mask describes the required logic functions on the array and then defines a unique interconnect pattern. Artwork for the mask is generated automatically from RCA's extensive library of logic macros (predefined logic structures or cells).

Automated GA software has proved to be very effective, providing 100% connectivity and gate utilization up to 98% for complex random logic applications.

### PaCMOS (Programmable Automated Cell CMOS) Standard Cells

PaCMOS is an RCA computer-automated design approach to LSI. It is based on a group of circuit building blocks, called standard cells, that can be automatically chosen, placed, and interconnected by a computer program.

RCA has a large and expanding library of previously designed, verified, and life-tested standard cells — available in C<sup>2</sup>L (closed-geometry silicon-gate CMOS-bulk), bulk CMOS (5μ and 3μ), and silicon-gate CMOS on sapphire substrate.

As a design aid in selecting appropriate cells, a standard cell notebook is available. In it, all of the logic cells are fully characterized and documented in the form of data sheets.

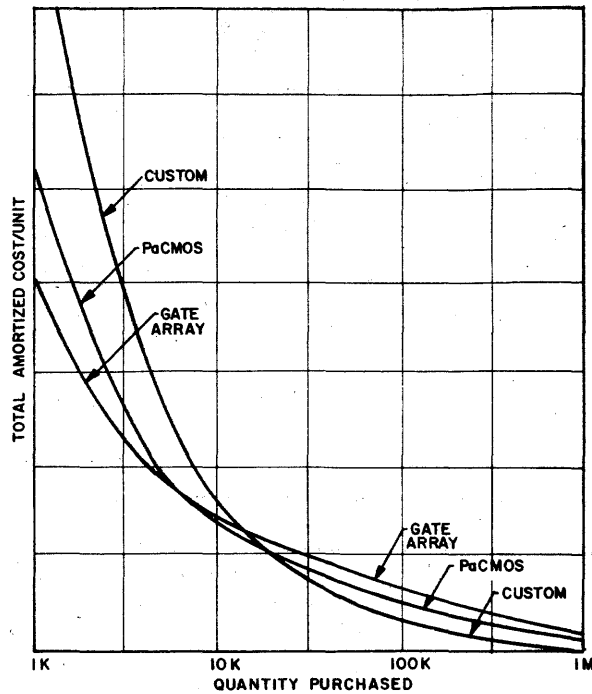
It provides the name and function of the cell, the circuit and logic configuration, the Boolean equation, the truth table, and the dynamic performance data.

By using the standard cell notebook, the designer can match his particular circuit configuration with the available standard cells and generate an input net list.

In addition to the standard cells, PaCMOS can also accommodate subchip options. Regular structures such as RAMs, ROMs, PLAs, PROMs, EPROMs, and register stacks, which have been remotely designed by conventional handcrafted techniques, can be automatically connected while the remainder of the host chip is computer generated.

#### PaCMOS VERSUS GA

A PaCMOS semicustom LSI offers attractive tradeoffs between development time, die size, and cost for many IC designs. Although its development time is slightly longer than a gate array, its die size is smaller. And in volume production PaCMOS can provide significant cost advances over GA designs.



92CS-35972

### Design Automated Tools

#### MIMIC

A powerful software simulation program, called MIMIC, allows designers of GAs, as well as other RCA semicustom and custom LSI devices, to model the logical operation of digital circuits before device fabrication. Through the program, designers can discover logical flaws, race, hazard, or spike conditions, and timing and critical path uncertainties.

#### CRITIC (Computer Recognition of Illegal Technology in Integrated Circuits)

Another software tool available in the design of GAs, as well as other RCA semicustom and custom LSI devices, is the CRITIC program. CRITIC checks mask artwork by describing the mask as a series of polygons in terms of the vertices that define them. The program then automatically searches for and reports on minimum tolerance violations (width, spacing, and enclosure) and illegal topology relations (overlap, abut, cross, contain, and disjoint).

#### AFTER

The AFTER program (Automatic Functional Test Encoding Routine) aids the IC designer in generating functional test patterns required for the testing of digital IC's on Automatic Test Equipment (ATE). Input/output truth tables, representing these tests, can be created by the designer using the MIMIC logic simulation program by specifying the IC's inputs to MIMIC; its outputs are calculated by MIMIC. ATEs currently supported are the Fairchild Sentry/Sentinel/ Series 20, the Teradyne J283, and the Datatron test systems.

#### CONCERT

(CONnectivity CERTification) is a layout analysis program which aids the verification of the logical and electrical correctness of the mask artwork produced by the APAR automatic layout programs. The layout generated by RCA's APAR programs (MP2D and GA) consists of standard cell placements and wire interconnects.

RCA

CUSTOM/SEMICUSTOM

# Custom and Semicustom LSI

## Features

- Full Range of Cost-Effective Design Capabilities
  - Gate Arrays — automatically placed and routed
  - CMOS I ( $5\mu$ ) — CMOS II ( $3\mu$ )
  - PaCMOS — standard cells
  - Full Custom LSI

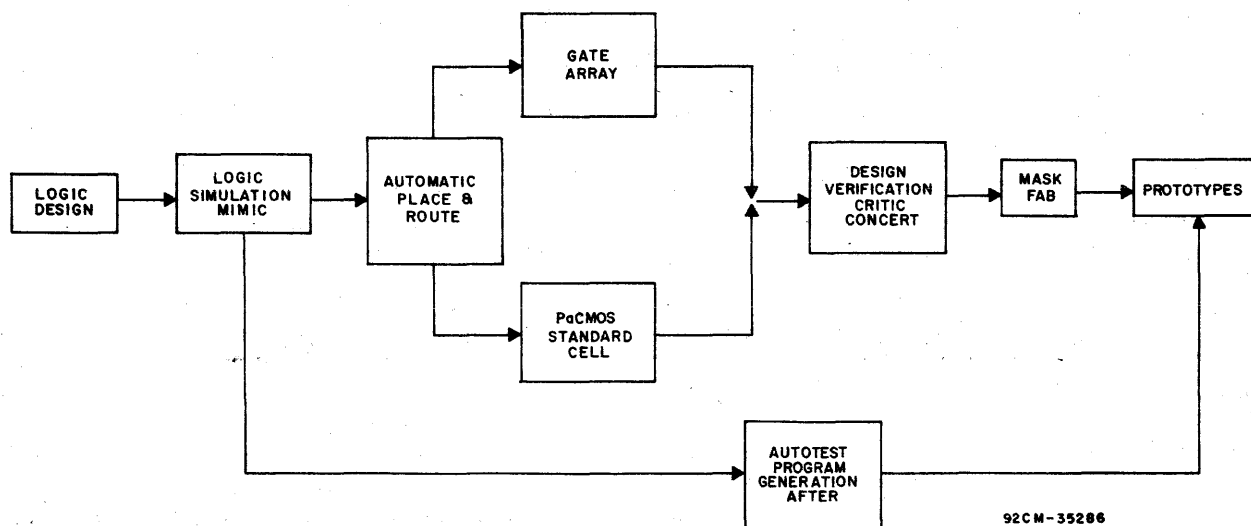
- Fully Supported Design Automation
  - MIMIC — logic simulation
  - CRITIC — design rule verification
  - AFTER — test program generation
  - CONCERT — connectivity check
- Long Experience
  - More than 15 years of custom/semicustom service to demanding customers.

## Comparison of Custom/Semicustom LSI Options\*

	Semicustom		Full Custom
	Gate Array	PaCMOS Std. Cell	
Development Cost	0.2	0.3	1
Development Time	0.17	0.23	1
Risk (1st time success)	1	1.05	2
Final Die Size	1.5-2.0	1.1-1.4	1
Production Die Cost	3-4	1.5-2.0	1
Efficiency of Chip Utilization	0.8	1	1

\*All comparisons normalized to a full custom basis assuming a 1000 gate complexity function.

## RCA Design Automation System



92CM-35286

RCA presently offers a variety of Gate Arrays classified according to:

Gate complexity: 250 to 6000 gates

RCA offers 17 Automated Universal Arrays (AUAs). The interconnect mask patterns for AUAs are generated solely by user-accessible software via the techniques of design automation (DA). AUAs are classified according to:

Process technology:

- CMOS I —  $5\mu$  silicon gate
- CMOS II —  $3\mu$  silicon gate
- CMOS/SOS —  $4\mu$  silicon on sapphire
- CMOS II — Double level metal —  $3\mu$  silicon gate

RCA presently offers a variety of PaCMOS cell libraries as follows:

Gate complexity: User defined

Process technology:

- CMOS silicon gate ( $C^2L$ )
- CMOS silicon on sapphire (CMOS/SOS)
- CMOS I —  $5\mu$  bulk CMOS
- CMOS II —  $3\mu$  bulk CMOS

RCA will soon offer additional APAR cell libraries. They are classified by:

Process technology:

- CMOS/SOS —  $2\mu$
- CMOS II — Double level metal

All of the above will be offered in plastic, ceramic, side-brazed ceramic, and leadless and leaded chip carrier packages.



## SP7000 SERIES CMOS GATE ARRAYS

### FEATURES

- Utilizes SPI's Selective Oxidation, Silicon-Gate CMOS Process.
- High Speed 3.5nS Gate Delay
- Low Cost
- Operates Over 3 to 7 Volt Supply Range
- Prototypes 8-10 Weeks
- TTL/CMOS Compatible Inputs

### SP7000 SERIES SPECIFICATIONS

Part Number	Number of Gates	Number of Pads	Number of Output Drivers	Die Size	Typical Output Current $I_{OL} @ V_{OL} = 0.4V$	Typical Output Current $I_{OL} @ V_{OH} = 4.2V$
SP7000	48	16	14	76 x 73	6mA	6mA
SP7001	80	20	16	86 x 93	10mA	8mA
SP7002	147	24	16	103 x 107	12mA	8mA
SP7003	300	40	22	118 x 104	6mA	1.5mA
SP7004	300	40	36	127 x 158	10mA	8mA
SP7005	544	56	32	132 x 150	6mA	1.5mA
SP7010	1000	76	40	185 x 185	6mA	1.5mA
SP7210	1200	84	72	186 x 251	6mA	4mA

Note: Some device specifications can be altered to meet customer requirements.

SPI's gate array in HCMOS will provide the key to reducing board space, lowering power requirements and improving cost effectiveness.

Whatever your gate array needs, SPI can tailor an interface program to match them. If you have a specific circuit you would like to implement in HCMOS, SPI will be glad to review your application and provide a detailed quotation on cost, delivery and technical parameters.

## CHLA CMOS LOGIC ARRAY FAMILY

### FEATURES

- Utilizes SPI's Selective Oxidation Silicon Gate
- CMOS Process
- Low static power dissipation —  $1.0 \mu W$
- High speed
- 3 to 6 volt power supply range
- Full TTL interface compatibility
- High noise immunity
- Low cost semi-custom.

### DESCRIPTION

The CHLA is a logic array with uncommitted logic cells driving output structures that may be configured as normal output buffers, 3-state buffers, or as D type edge triggered latches. To design with these devices the designer needs only to supply the

necessary logic equations. Depending on the logic equations up to 64 product terms may be implemented. Since all logic gates are CMOS the static power supply current is independent of the number of product terms and is less than  $2.0 \mu A$ .

Part Number	Description		I/O
	Inputs	Outputs	
SP810H8	10	8	
SP812H6	12	6	
SP814H4	14	4	
SP818H2	16	2	
SP816C1	16	2	
SP810L8	10	8	
SP812L6	12	6	
SP814L4	14	4	
SP816L2	16	2	
SP816L8	10	8	6*
SP816R8	8	8	8**
SP816R6	8	6	8**
SP816R4	8	4	8**

\* = 3-State  
\*\* = Latched 3-State

# C-MOS HARD ARRAY LOGIC SI-HAL SERIES 20 & 24

## DESCRIPTION

The Si-HAL family provides a rapid economical way of customizing several SSI or MSI integrated circuits into one device. Si-HALs can be used to replace bipolar PALs and HALs or they can be developed from a unique logic schematic. Unlike bipolar HALs, C-MOS Si-HALs are not limited to fixed logic functions. You have a totally uncommitted logic format to work with in developing your semi-custom circuit. You may choose from a wide range of logic functions including, but not limited to, AND, NOR, OR, NAND gates and various flip flops.

All of the Monolithic Memories HAL Series 20 & 24 can be pin for pin replaced with a Si-HAL Series 20 & 24.

The C-MOS Si-HAL family features an advanced silicon-gate process which gives all the advantages of C-MOS, and still is a direct replacement for many bipolar PALs and HALs.

Si-Fab II is a unique process, which goes beyond the performance characteristics of conventional silicon-gate C-MOS. It has the same low power characteristics, but in addition demonstrates improved speed. On-chip gate delays are comparable to silicon-gate (1-3ns), but the data entry speeds are improved 50-75%. This speed enhancement is made possible through our unique Si-Fab II output drivers. Also, output drive is improved to handle up to 20 LSTTL loads.

## HAL DEVELOPMENT

The following procedures can be followed in developing a Si-HAL semi-custom circuit.

1. Purchase a bipolar PAL and debug it to your circuit requirement.
2. Send Si-Fab two of your PAL circuits with either
  - a. copy of circuit logic equations
  - or
  - b. a schematic drawing (checked for revisions and accuracy)

**NOTE:** If you are not interested in developing a circuit using a PAL first, a Si-HAL can be directly developed from your circuit schematic diagram.

**Call Si-Fab with any questions you may have regarding Si-HALs or semi-custom circuit development**

## MAIN BENEFITS

- Semi-custom solution to random logic
- Direct C-MOS replacement of bipolar HALs and PALs
- Easy, economical implementation from logic schematics
- Si-Fab II C-MOS provides low power, high speeds, high output drive
- Variety of packages available

## PART NO. DESCRIPTION

PART NO.	DESCRIPTION
Si-HAL 10H8	Octal 10input And-Or Gate Array
Si-HAL 12H6	Hex 12input And-Or Gate Array
Si-HAL 14H4	Quad 14input And-Or Gate Array
Si-HAL 16H2	Dual 16input And-Or Gate Array
Si-HAL 16C1	16input And-Or/And-Or-Invert Gate Array
Si-HAL 10L8	Octal 10input And-Or-Invert Gate Array
Si-HAL 12L6	Hex 12input And-Or-Invert Gate Array
Si-HAL 14L4	Quad 14input And-Or-Invert Gate Array
Si-HAL 16L2	Dual 16input And-Or-Invert Gate Array
Si-HAL 16L8	Octal 16input And-Or-Invert Gate Array
Si-HAL 16R8	Octal 16input Registered And-Or Gate Array
Si-HAL 16R6	Hex 16input Registered And-Or Gate Array
Si-HAL 16R4	Quad 16input Registered And-Or Gate Array
Si-HAL 16X4	Quad 16input Registered And-Or-Xor Gate Array
Si-HAL 16A4	Quad 16input Reg And-Carry-Or-Xor Gate

## PACKAGING

Si-HALs can be shipped in chip or package form. A variety of packages are available. Request of quote for your package requirements.

Series 20 has 20 lead maximum.  
Series 24 has 24 lead maximum.

## PRICING

Prices will vary depending on circuit complexity, quantity, and package required. A one time development charge will be assessed on all orders, but this will be partially rebated on production orders according to quantity on all orders exceeding 1000 pieces.

Price List available on request.



Si-Fab Corporation • 27 Janis Way • Scotts Valley, CA 95066 • (408)438-6800

# ABBREVIATIONS OF COMPANY NAMES

**Action Ins** Action Instruments  
**AD** Analog Devices  
**ADT** Advanced Digital Technology  
**Advent** Advent Products, Inc.  
**Alphatron** Alphatron  
**AMA** American Automation  
**AMD** Advanced Micro Devices  
**AMI** American Microsystems, Inc.  
**Amperex** Amperex Electronic Corp.  
**Analogic** Analogic  
**Analog Sys** Analog Systems  
**APC** Applied Micro Circuits  
**Apex** Apex Microtechnology  
**APM** Applied Microsystems Corp.  
**Appl Sys** Applied Systems Corp.  
**APT** Applied Microtechnology  
**Aptek** Aptek Microsystems  
**Array Tech** Array Technology  
**AWI** AWI Electronics

**Barvon** Barvon Research  
**Bedford** Bedford Computer Systems Inc.  
**Burr-Brown** Burr-Brown

**CAE** Computer Aided Engineering  
**Cal Devices** California Devices  
**Cermetek** Cermetek  
**CGRS** CGRS Microtech Inc.  
**Cherry** Cherry Semiconductor  
**CIC** Custom Integrated Circuits  
**CirTech** Circuit Technology  
**Citel** Citel  
**Comlinear** Comlinear Corporation  
**CMA** Custom MOS Arrays  
**Comark** Comark Corp.  
**Comdial** Comdial Semiconductor  
**Comp Auto** Computer Automation  
**Compas** Compas Microsystems  
**Cont Logic** Control Logic Inc.  
**Control Sys** Control Systems Microsystems Div.  
**CreMicro** Creative Micro Systems  
**Cromemco** Cromemco, Inc.  
**Cubit** Cubit Inc.  
**Curtis** Curtis Electro Devices, Inc.  
**Cybernetic** Cybernetic Micro Systems  
**Cybersys** Cybersystems  
**Cybertek** Cybertek Inc.

**Data General** Data General  
**Data I/O** Data I/O  
**Data Trans** Data Translation  
**Datel** Datel  
**Datricon** Datricon Corporation  
**DDC** Data Devices Corporation  
**DEC** Digital Equipment Corporation  
**Die-Tech** Die-Tech  
**Digelec** Digelec Corp.  
**Digitek** Digitek, Inc.  
**Dionics** Dionics Inc.  
**Dist Comp** Distributed Computer Systems  
**Divers Tech** Diversified Technology

**E-HI** E-H International, Inc.  
**EDI** Electronic Designs Inc.  
**Elind** Elind Electronica Industriale  
**EMM-SESCO** EEM-SESCO  
**Emulogic** Emulogic Inc.  
**ETI Micro** ETI Micro  
**Exar** Exar Integrated Systems  
**Exel** Exel Microelectronics

**Fairchild** Fairchild  
**Ferranti** Ferranti Electric  
**Force** Force Computers  
**Fujitsu A** Fujitsu America  
**Fujitsu** Fujitsu Microelectronics, Inc.

**GI** General Instrument  
**GTE Micro** GTE Microcircuits

**Harris** Harris Semiconductor  
**Heurikon** Heurikon Corp.  
**Hilevel** Hilevel Technology, Inc.  
**Hitachi** Hitachi America, Ltd.  
**Holt** Holt Inc.  
**HP** Hewlett-Packard  
**Hughes** Hughes Aircraft, Solid State Products  
**Hybrid Sys** Hybrid Systems  
**HyComp** HyComp

**ICC** International Cybernetics  
**IDT** Integrated Device Technology  
**IMI** International Microcircuits, Inc.  
**IMP** International Microelectronic Products

**IMS** Industrial MicroSystems Inc.  
**Infosphere** Infosphere  
**Inmos** Inmos  
**IntCirEng** Integrated Circuit Engineering  
**IntCirSys** Integrated Circuit Systems  
**IntCompSys** Integrated Computer Systems  
**Int Tech** Integrated Technology Corp.  
**Intech** Intech Microcircuits Div.  
**Intel** Intel  
**Interdesign** Interdesign  
**Intersil** Intersil  
**Intronics** Intronics  
**ITT** ITT Semiconductors

**Kinetic Sys** Kinetic Systems  
**Kontron** Kontron Electronics

**Lambda** Lambda Semiconductor  
**Linear Tech** Linear Technology  
**LSI Comp** LSI Computer Systems  
**LSI Logic** LSI Logic Corporation

**Master Logic** Master Logic Corporation  
**Matrix** Matrix Corp.  
**Matrox** Matrox Electronic Systems  
**Maxim** Maxim Integrated Products  
**MCC** Micro-Computer Control  
**MCE** MCE Electronics  
**Micrel** Micrel  
**Micro Innov** Micro Innovators  
**Micropac** Micropac Industries  
**Micro Net** Micro Networks  
**Micro Pwr** Micro Power Systems  
**Micro Sci** Micro Sciences Corp.  
**Micro Tech** Microcircuits Technology  
**Micro-Link** Micro-Link Corporation  
**Micron** Micron Technology  
**MillerTron** MillerTronics  
**Miller** Miller Technology  
**Mitel** Mitel Semiconductor  
**Mitsubishi** Mitsubishi Electronics  
**MMI** Monolithic Memories, Inc.

**Mostek** Monolithic Systems Corp.  
**Motorola** Mostek  
**MRC** Motorola Semiconductor  
**Murray** MRC Systems  
**Monosil** Murray Consulting

**National** National Semiconductor  
**NCM** NCM Corp.  
**NCR** NCR Corp., Microelectronics Division  
**NEC** NEC Electronics  
**Nitron** Nitron

**OAE** Oliver Advanced Engineering  
**Octagon** Octagon Systems Corp.  
**OEI** Optical Electronics Inc.  
**Ohio Sci** Ohio Scientific  
**OKI** OKI Semiconductor  
**Omnibyte** Omnibyte Corp.  
**Onset** Onset Computer Corp.

**Panasonic** Panasonic  
**Pico Design** Pico Design  
**Polycore** Polycore Electronics  
**Plessey** Plessey Semiconductors  
**PMI** Precision Monolithics, Inc.  
**PragDes** Pragmatic Design Inc.  
**Pro-Log** Pro-Log Corp.

**Quay** Quay Corp.

**Raytheon** Raytheon Semiconductor  
**RCA** RCA Solid State Division  
**RCI Data** RCI Data  
**RELMS** Relational Memory Systems  
**Reticon** Reticon  
**RIFA** RIFA  
**Rockwell** Rockwell, Microelectronic Devices  
**RTC** Riehl Time Corporation

**Sanyo** Sanyo  
**SBE** SBE, Inc.  
**SEEQ** SEEQ Technology, Inc.  
**SPI** Semi Processes Inc.  
**Siemens** Siemens  
**Si-Fab** Si-Fab  
**Signetics** Signetics  
**SGS** SGS Semiconductor  
**Sharp** Sharp  
**Silicon G** Silicon General  
**Siliconix** Siliconix  
**Silicon Sys** Silicon Systems Inc.  
**Siltronics** Siltronics  
**SMC** Standard Microsystems Corp.  
**S MOS** S MOS Systems  
**Solarise** Solarise Enterprises  
**Solitron** Solitron Devices  
**Sprague** Sprague Electric Company  
**SSM** Solid State Micro Technology for Music  
**SSS** Solid State Scientific  
**Stag** Stag Microsystems  
**STC** Storage Technology Corp.  
**STD** STD Microsystems  
**Struc Des** Structured Design Inc.  
**Stynetic** Stynetic Systems  
**Sunrise** Sunrise Electronics  
**Sunshine** Sunshine Semiconductor  
**Supertex** Supertex Inc.  
**Symtek** Symtek Corp.  
**Synertek** Synertek  
**Sys Innov** Systems Innovations

**Tau Zero** Tau Zero Inc.  
**Technitrol** Technitrol  
**Tektronix** Tektronix  
**Teledyne C** Teledyne Crystalonics  
**Teledyne P** Teledyne Philbrick  
**Teledyne S** Teledyne Semiconductor  
**Telefunken** Telefunken  
**Telmos** Telmos  
**Teltone** Teltone Corporation  
**TI** Texas Instruments  
**Third Domain** Third Domain  
**Thomson-CSF** Thomson-CSF Components Corp.  
**Toshiba** Toshiba America  
**Trans-Data** Trans-Data  
**TRW** TRW-LSI Products

**Unitrode** Unitrode  
**Universal** Universal Semiconductor, Inc.

**Varix** Varix Corp.  
**VLSI Design** VLSI Design Associates  
**VTI** VLSI Technology, Inc.  
**Votrax** Votrax

**Weitek** Weitek Corporation  
**Western** Western Digital  
**Wintek** Wintek Corp.

**Xicor** Xicor, Inc.  
**Xycom** Xycom

**Zendex** Zendex Corp.  
**Zilog** Zilog  
**ZyMOS** ZyMOS Corporation  
**Zytrex** Zytrex Corp.

# Signetics

## BIPOLAR LSI Gate Array/Semi-Custom Products

### SEMI-CUSTOM SERVICE

#### FLEXX™ ARRAY

8A1200

CG1001

8A1260

8A1542

8A1664

8A1864

8A2176

CCL

ACE0600/0900

ACE 1320/1400/2200

SCC 0330-M, SCC 0450-M, SCC 0700-M, SCC 1100-M

SCC 0335-H, SCC 0455-H, SCC 0705-H, SCC 1105-H

SCC 1250-SH, SCC 2500-SH, SCC 5000-SH

## signetics

a subsidiary of U.S. Philips Corporation

Signetics Corporation  
811 E. Arques Avenue  
P.O. Box 3409  
Sunnyvale, California 94088-3409  
Telephone 408/739-7700

Signetics

CUSTOM/SEMICUSTOM

## SEMICUSTOM LSI

## SEMICUSTOM SERVICE

### THE COMPLETE SEMICUSTOM SOLUTION

To meet your semicustom needs Signetics has developed a comprehensive semicustom service that offers one-stop shopping with state-of-the-art technology. This service gives you a choice of silicon systems ranging from CMOS to T<sup>2</sup>L and ECL, the widest range of semicustom products in the industry, with performance capable of meeting all your semicustom logic needs. All are supported with complete in-house capability for processing and fabrication, will full service CAD, which simplifies your design task and guarantees first-pass success for your semicustom program.

### STATE-OF-THE-ART CAD

Signetics' CAD system encompasses the complete semicustom design process, from schematic input through auto place and route, design rule checking and test generation. This fully automated procedure speeds design and makes your job easy and guarantees that your semicustom chip is developed without error and on schedule. With Signetics' CAD you are using the most up-to-date semicustom design system available.

### SCHEMATIC INPUT — FAST AND EASY

In designing a semicustom chip the first step is to enter your circuit into the computer. The most advanced method of circuit entry is Signetics schematic input system. This system only requires you to enter your schematic by use of a keyboard and mouse, which is fast and easy. The computer then automatically converts your schematic to a net list which is used for simulation and auto place and route.

For circuit entry Signetics gives you a choice of two schematic input systems, one with software originally developed by Future Net and one by Mentor. The Future Net system operates on a low cost IBM Personal Computer which you can use in your own shop or in a Signetics' sales office. The Mentor schematic input system operates on an Apollo computer which is available in one of Signetics' design centers. With either input system you will find the schematic entry step fast and easy.

### COMPUTER SIMULATION — EASY AND SURE

Once your schematic has been entered, the next and final step is to simulate the logic using our TEGAS 5 Program. To do this you enter your test vectors then submit the job from your IBM PC to Signetics' computer via telephone link. If you are using one of our design centers the simulation can be performed on-site with the Apollo computer. Upon completion of simulation the computer returns to you the resulting output vectors to allow you to check your design. Computer simulation ensures that your circuit is functioning properly before Signetics manufactures your semicustom chip.

### WHAT SIGNETICS DOES — MORE CAD

After completion of simulation your job is finished. Signetics takes over and completes the manufacture of your semicustom device, which involves the use of additional

Signetics CAD. The effects of wire delay, if any, are compensated using our WIDGET program. Using your net list which was produced by the schematic input system, Signetics performs a comprehensive design rule check, then routes the chip automatically. Using your test vectors Signetics automatically produces a test tape for final testing using our SENGEN program. The routed chip is then masked and processed using our advanced semiconductor methodology.

In performing Signetics part of the task, a key step is auto place and route, and Signetics has placed a high emphasis on developing these programs. The MEDS program automatically routes at the gate level and is used for gate arrays and creation of macros. The CALMP program is used for more advanced structures, such as Signetics FLEXX<sup>TM</sup> Array. CALMP assembles macros and gates with automatically variable routing space for optimal packing of the chip. These and other similar programs allow Signetics to offer comprehensive use of auto-routing in all our semicustom products.

### FULL IN-HOUSE CAPABILITY

As a full capability semiconductor supplier and in semicustom since 1975, Signetics has all the in-house support necessary for producing your semicustom chip. In addition to a full service CAD facility, Signetics has complete masking, wafer fabrication and processing, including military processing, for Source Control Drawing or MIL STD 883B parts production. This full capability ensures availability of parts and that your semicustom device will be processed according to your needs and on schedule.

### A FULL CHOICE OF SILICON

With Signetics you can implement your semicustom requirement with silicon that is advanced in technology and architecture. In technology you have a choice of three levels of CMOS and five kinds of bipolar arrays with gate speed ranging from 8 nsec to 0.5 nsec and various output drive levels. In architecture these products are structured as gate arrays, masterslice, composite cell, and FLEXX<sup>TM</sup> arrays, with up to 5000 gates and 128 I/Os. These devices have been configured to encompass the complete range of semicustom applications to allow Signetics to meet your own specific needs.

### CMOS GATE ARRAYS

Signetics CMOS gate arrays are oxide-isolated, silicon gate devices built on an epi substrate which virtually eliminates latch-up. Input/Output is compatible with CMOS or LSTTL logic. These are highly advanced CMOS gate arrays with a full selection of technology and architecture.

**M-Series CMOS Gate Arrays** Single layer metal, 330 to 1100 gates, up to 68 I/Os, with 8 nsec gate delay typical for a 2 output gate with 5V V<sub>DD</sub>.

## SEMICUSTOM LSI

## SEMICUSTOM SERVICE

**H-Series CMOS Gate Arrays** Single layer metal, 330 to 1100 gates, up to 68 I/Os, with 4 nsec gate delay typical for a 2 output gate with 5V  $V_{DD}$ .

**SH-Series CMOS Gate Arrays** Dual layer metal, 1250 to 5000 gates, up to 128 I/Os, with 2 nsec gate delay typical for a 2 output gate with 5V  $V_{DD}$ .

### BIPOLAR SEMICUSTOM DEVICES

Signetics bipolar semicustom devices are available with both junction isolation and oxide isolation, for T<sup>2</sup>L and ECL applications. They are designed for high speed and high output drive.

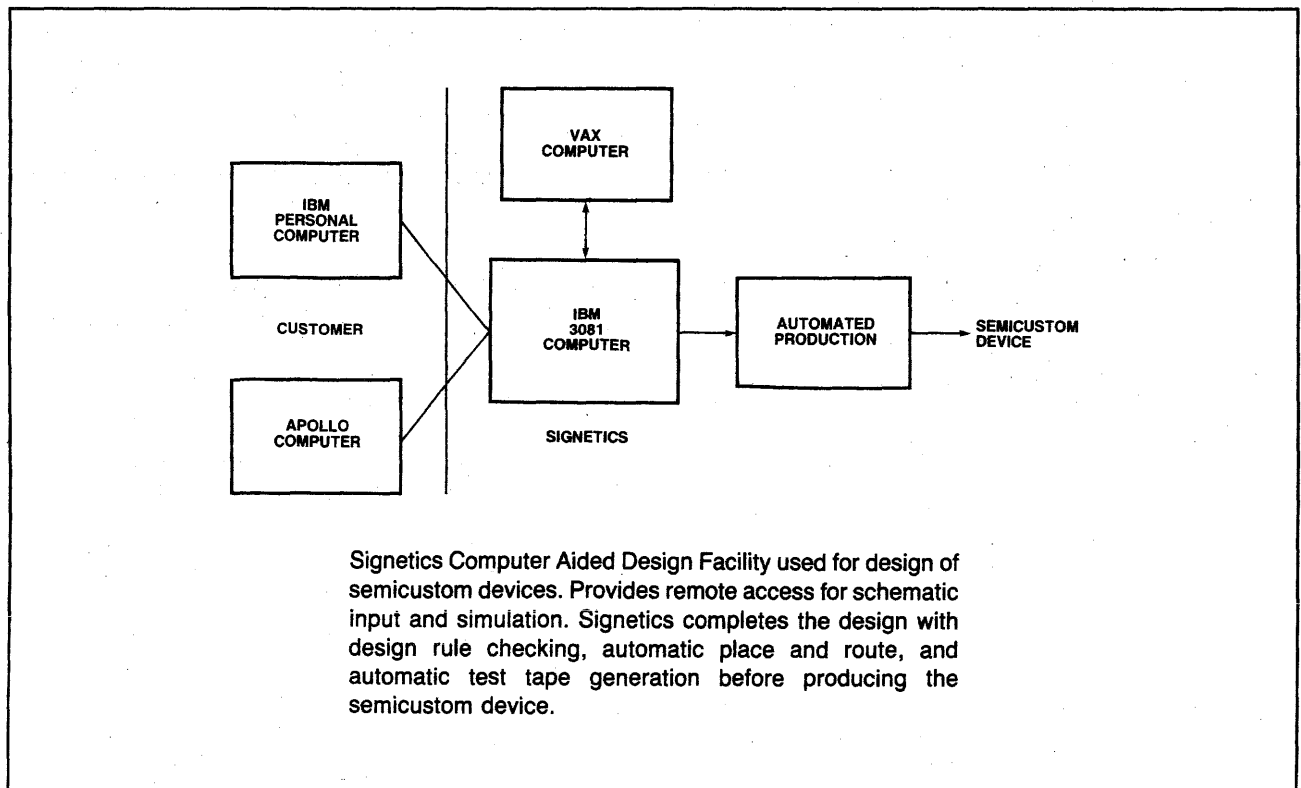
**8AXXXX Series ISL Gate Arrays** Dual layer metal junction isolated integrated Schottky Logic for T<sup>2</sup>L applications, 1200 to 2100 gates, up to 76 I/Os, with 4 nsec gate speed typical. Full military criteria.

**Composite Cell Logic** Three standard cell dual-layer metal libraries ranging from 3.5 nsec to 5 nsec typical gate delay. Up to 1000 actual gates, I/Os limited by package. Up to 80 mA output drive for T<sup>2</sup>L applications. Full military criteria.

**8HXXXX Series ISL Gate Arrays** Dual layer metal, oxide isolated Integrated Schottky Logic for T<sup>2</sup>L applications, 1600 to 3200 gates, up to 120 I/Os, with 1.5 nsec gate speed typical. Full military criteria.

**FLEXX™ Array** User-created or standard cells, dual-layer metal, variable routing space for optimal chip size, up to 2000 gates, I/Os limited by package. Oxide-isolated Integrated Schottky Logic for T<sup>2</sup>L applications, 1.5 nsec gate delay typical. Full military criteria.

**ACE Masterslice Array** Oxide isolated CML for ECL 10 K, ECL 100 K and T<sup>2</sup>L applications, 600 to 2200 gates, optional RAM on chip, up to 128 I/Os, 0.5 nsec gate delay, air cooled.



## SEMICUSTOM LSI

## FLEXX™ ARRAY

### FEATURES

- 1.5 nsec typical gate delay
- Up to 2000 actual gates
- I/Os limited by package
- 250uW per gate
- Up to 48 mA output drive
- TTL Compatible
- Full CAD support
- Automatic schematic input
- Circuit simulation
- Automatic place and route
- Automatic macro generation
- Hard and soft macros
- Variable die size
- Plastic and ceramic chip carriers
- Standard DIPs

### BENEFITS

- Fast, easy design
- Efficient use of silicon
- Proprietary LSI device
- High speed
- Replaces up to 100 SSI/MSI parts
- Reduces PCB area
- Saves manufacturing costs
- Reduces size, weight and power
- Improves system reliability

### PRODUCT DESCRIPTION

The FLEXX™ array represents a major advance in semicustom technology which combines a new concept in architecture with the most advanced CAD available. With it you can create your proprietary semicustom LSI device quickly and easily, much like gate array or standard cell methodology, but with silicon utilization as efficient as standard LSI products designed by traditional handpacking methods. As a result of its superior design and silicon utilization efficiency the FLEXX™ Array is being used by Signetics to develop standard LSI products. The FLEXX™ Array is now available to Signetics customers as a semicustom development tool.

#### FLEXX™ Array Architecture

As shown in the figure, the FLEXX™ Array is composed of macros, which are rectangular assemblies of gates, with additional random gates included as required. The macros are variable in length and width for optimum routing efficiency, and the routing channel width is variable to accommodate only those traces required to route the chip. No unused gates are included. As a result of this architecture, the FLEXX™ Array is much more efficient in silicon utilization than a gate array or standard cell array, which means that the end device will have a smaller die and therefore a lower cost.

#### Multi-Level Software

To assemble the FLEXX™ Array Signetics uses multi-level automatic place and route software. The first level automatically generates the macros; it establishes the macro width and interconnects the gates within the macro. The second level places and interconnects the macros as well as individual gates; it varies the spacing between macros to accommodate the required interconnecting traces. This process is fully automatic and can be done quickly which speeds the development of your FLEXX™ Array.

#### Fully Automatic Design

To design a FLEXX™ Array the user enters the schematic and test vectors with a remote terminal then reviews the computer simulation of the circuit. In this procedure the user has the choice of using established ("hard") macros from Signetics computer library, modifying these macros or creating new ("soft") macros as required. Once the simulation is completed, Signetics takes over and routes the chip and procedures prototypes within ten weeks.

#### Silicon Technology — Performance

The FLEXX™ Array is a methodology for assembling a logic design on silicon; it is therefore largely independent of the particular silicon technology and can be used with CMOS and varieties of bipolar technology.

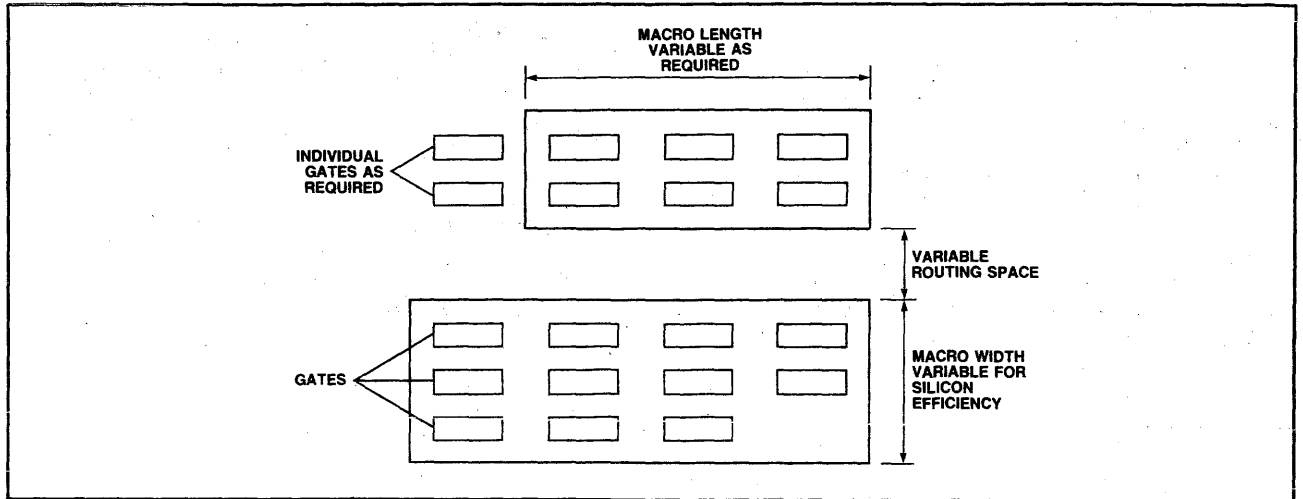
The first technology used to implement the FLEXX™ Array is oxide isolated ISL or Integrated Schottky Logic. ISL provides a superior speed power product with the speed of LSTTL at one-tenth the power. With oxide isolation, both the speed-power product and the gate delay of ISL are reduced by a factor of three, giving the FLEXX™ Array speed approaching ECL and power efficiency that allows large scale integration in a plastic package without special cooling.

With oxide-isolated ISL the gate used in the FLEXX™ Array has a typical delay time of 1.5 nsec. Output buffers are capable of driving up to 48 mA. The number of actual gates which can be placed on the FLEXX™ Array is limited to about 2000 by thermal constraints. The number of I/Os is limited only by package pin-out. Available packages currently include standard plastic and ceramic DIPs and a variety of chip carriers.

**SEMICUSTOM LSI**

**FLEXX™ ARRAY**

**FLEXX™ ARRAY ARCHITECTURE**





**ISL GATE ARRAY**

**8A1200**

**PRODUCT DESCRIPTION**

The 8A1200 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2), Schottky buffers (Figure 3) and LSTTL-compatible I/O cells (Figure 4). Thus, up to 1200 gates can be custom interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A1200 array is based on a technological subset of LSI called ISL (Integrated Schottky Logic). ISL combines the best features of low-power Schottky and I<sup>2</sup>L Bipolar technologies.

Designing with the 8A1200 is easy and fast, requiring no more than conventional logic design, logic simulation, and custom coding of metal interconnections among preprocessed logic gates on the array—refer to Table 1 for a comparison of ISL and 74LS logic functions. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting 1144 ISL NAND gates, using two layers of metal routing. Fifty-two Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 36 LSTTL I/O buffers can be specified. As shown in Figure 4, each I/O can be configured to implement any one of 11 different functions: inputs, input/output, totem-pole, open collector, and three-state.

**FEATURES**

- Customer programmable LSI
- 1144 ISL (NAND) gates
- Two-layer metal interconnection
- 52 Schottky buffers
- 36 I/O buffers
- LSTTL compatible
- Standard PNP inputs
- 8mA output current sink
- -55°C to +125°C ambient temperature
- 4ns gate speed (typical)
- Speed-power product—0.7 picojoules
- 22, 28, 40, or 44-pin package

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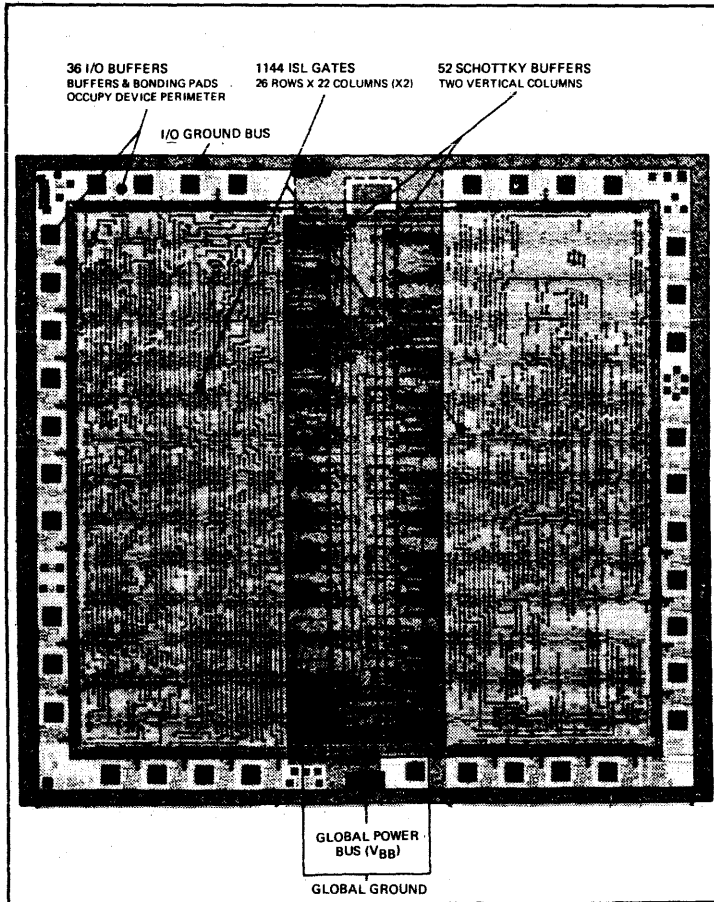


Figure 1. Internal Configuration of 8A1200 ISL Gate Array

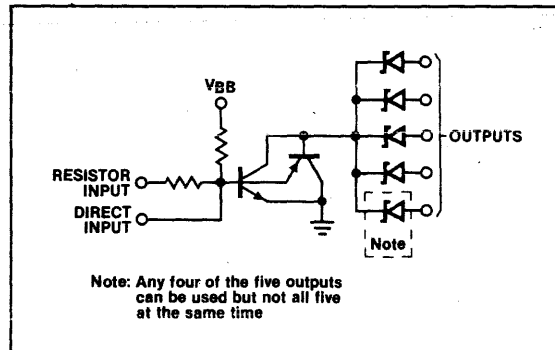


Figure 2. ISL Gate—Schematic Diagram

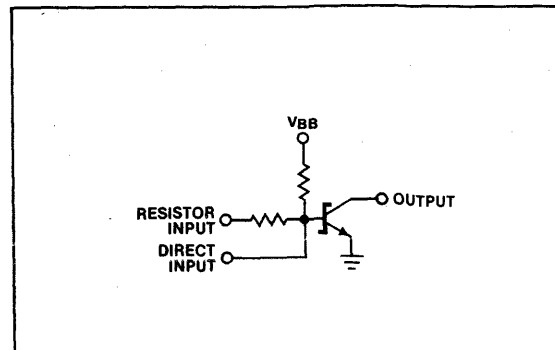
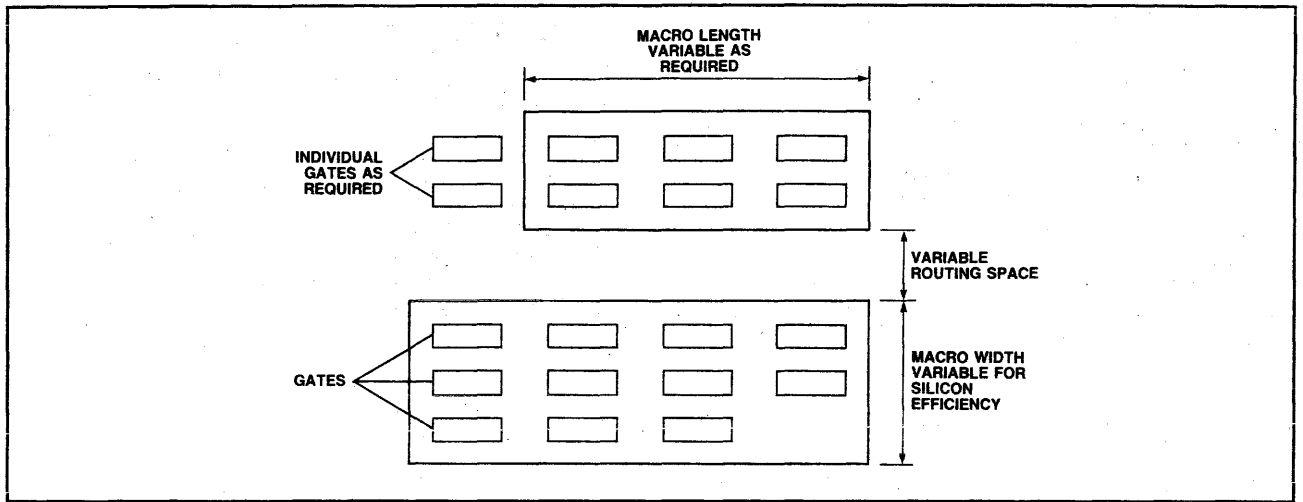


Figure 3. Schottky Buffer—Schematic Diagram

SEMICUSTOM LSI

FLEXX™ ARRAY

FLEXX™ ARRAY ARCHITECTURE



**ISL GATE ARRAY**

**8A1200**

**PRODUCT DESCRIPTION**

The 8A1200 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2), Schottky buffers (Figure 3) and LSTTL-compatible I/O cells (Figure 4). Thus, up to 1200 gates can be custom interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A1200 array is based on a technological subset of LSI called ISL (Integrated Schottky Logic). ISL combines the best features of low-power Schottky and I<sup>2</sup>L Bipolar technologies.

Designing with the 8A1200 is easy and fast, requiring no more than conventional logic design, logic simulation, and custom coding of metal interconnections among preprocessed logic gates on the array—refer to Table 1 for a comparison of ISL and 74LS logic functions. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting 1144 ISL NAND gates, using two layers of metal routing. Fifty-two Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 36 LSTTL I/O buffers can be specified. As shown in Figure 4, each I/O can be configured to implement any one of 11 different functions: inputs, input/output, totem-pole, open collector, and three-state.

**FEATURES**

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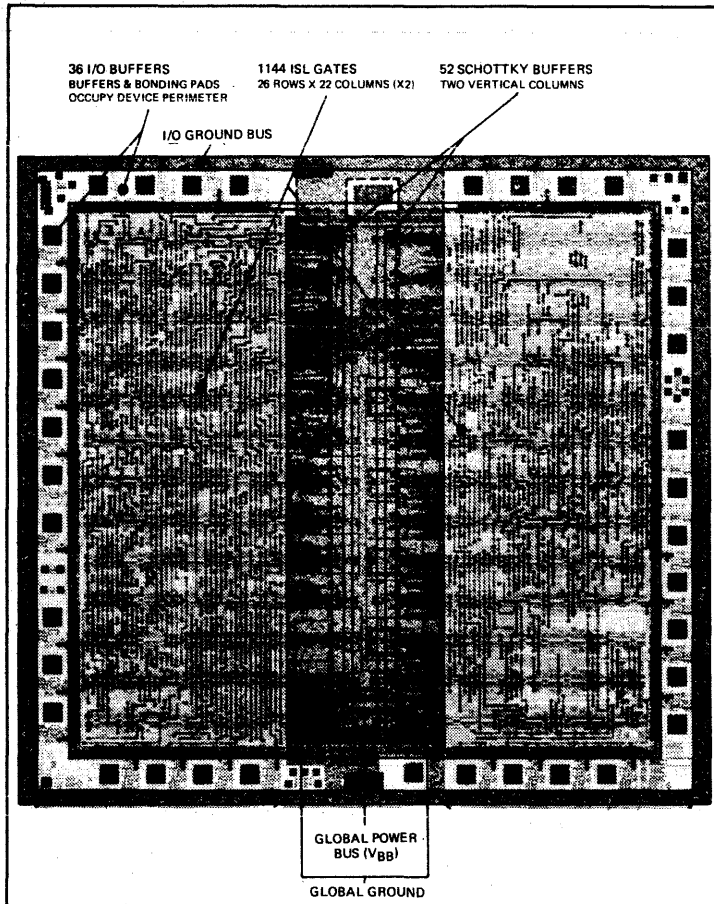


Figure 1. Internal Configuration of 8A1200 ISL Gate Array

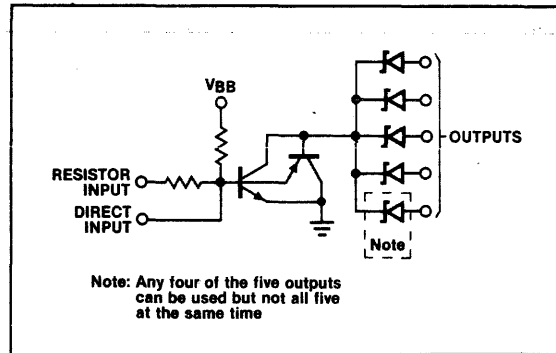


Figure 2. ISL Gate—Schematic Diagram

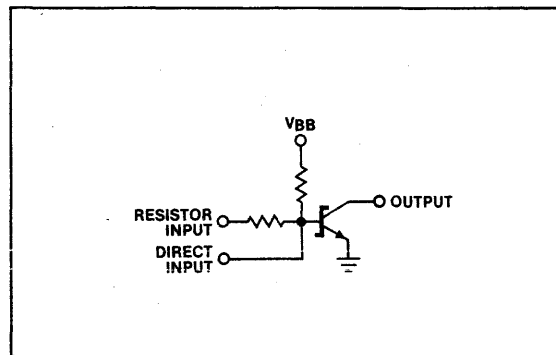


Figure 3. Schottky Buffer—Schematic Diagram

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# ISL GATE ARRAY

8A1200

Table 1. Comparison of ISL with 74LS Functions

LOGIC FUNCTION	PARAMETERS <sup>3</sup>	ISL <sup>1</sup>	74LS	LOGIC FUNCTION	PARAMETERS <sup>3</sup>	ISL <sup>1</sup>	74LS
NAND (7400)	Power (in mW)	0.30	2.00	D FLIP-FLOP (7474) C→Q	Power (in mW)	1.50	20.00
	T <sub>ON</sub> (in ns)	2.00	15.00		T <sub>ON</sub> (in ns)	28.00	40.00
	T <sub>OFF</sub> (in ns)	10.00	15.00		T <sub>OFF</sub> (in ns)	28.00	40.00
AND (7408)	Power (in mW)	0	11.00	D LATCH (7475) DATA→Q	Power (in mW)	1.20	30.00
	T <sub>ON</sub> (in ns)	0	20.00		T <sub>ON</sub> (in ns)	5.00	17.00
	T <sub>OFF</sub> (in ns)	2 <sup>2</sup>	15.00		T <sub>OFF</sub> (in ns)	12.00	17.00
EXCLUSIVE OR (7486)	Power (in mW)	1.20	12.50	4-INPUT MUX (74153) DATA→Q	Power (in mW)	1.50	25.00
	T <sub>ON</sub> (in ns)	18.00	22.00		T <sub>ON</sub> (in ns)	7.00	26.00
	T <sub>OFF</sub> (in ns)	24.00	30.00		T <sub>OFF</sub> (in ns)	16.00	26.00
EXCLUSIVE NOR (74266)	Power (in mW)	0.90	18.00	Notes: 1. Power and delay times are given for 150°C MAX. 2. T <sub>OFF</sub> is 2ns for each input; T <sub>OFF</sub> can be reduced to 0ns with a pullup cell which uses 0.3 mW. 3. LS power dissipation is based on V <sub>CC</sub> × I <sub>MAX</sub> .			
	T <sub>ON</sub> (in ns)	15.00	30.00				
	T <sub>OFF</sub> (in ns)	16.00	30.00				

## AC AND DC ELECTRICAL CHARACTERISTICS

<b>Conditions:</b>	Commercial—	Military—
	V <sub>CC</sub> = 5.0V (± 5%)	V <sub>CC</sub> = 5.0V (± 10%)
	V <sub>BB</sub> = 1.5V (± 10%)	V <sub>BB</sub> = 1.5V (± 10%)
	T <sub>A</sub> <sup>1</sup> = 0°C to 70°C	T <sub>A</sub> <sup>1</sup> = -55°C to 125°C

## ABSOLUTE MAXIMUM RATINGS

PARAMETER	DESCRIPTION	RATING	UNIT	PARAMETER	DESCRIPTION	RATING	UNIT
V <sub>CC</sub>	Supply voltage	+ 7.0	V	V <sub>O</sub>	Voltage applied to open-collector output in off-state	-0.5 to +7.0	V
V <sub>BB</sub>	ISL gate supply voltage	+ 7.0	V	T <sub>A</sub>	Ambient temperature, operating	-55 to +125	°C
E <sub>IN</sub>	Input voltage, continuous	-0.5 to +5.5	V	T <sub>STG</sub>	Storage temperature	-65 to +150	°C
I <sub>IN</sub>	Input current, continuous	-30 to +1.0	mA				

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
<b>ISL GATE (Internal)</b>									
I <sub>BB/G</sub>	Power supply current per gate			190			190	μA	
ILF	Input load factor			1			1	Unit load	
FO	Fanout			4			4	Unit load	
t <sub>pdAV</sub>	Average gate propagation delay $t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = one (1) ISL gate or Schottky buffer Fan-out = one (1) ISL gate or Schottky buffer		4	6		4	6	ns
t <sub>pdHL2</sub>	High-to-low propagation delay	Delay is inferred from circuit simulation		1	2		1	2	ns
t <sub>pdLH2</sub>	Low-to-high propagation delay			7	10		7	10	ns
<b>SCHOTTKY BUFFER (Internal)</b>									
I <sub>BB/G</sub>	Power supply current per gate			190			190	μA	
ILF	Input load factor			1			1	Unit load	
FO	Fanout			10			10	Unit load	
t <sub>pdAV</sub>	Average gate propagation delay $t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = one (1) ISL gate or Schottky buffer Fan-out = one (1) ISL gate or Schottky buffer		4	6		4	6	ns
t <sub>pdHL2</sub>	High-to-low propagation delay	Delay is inferred from circuit simulation.		1	2		1	2	ns
t <sub>pdLH2</sub>	Low-to-high propagation delay			7	10		7	10	ns

**ISL GATE ARRAY**

**8A1200**

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
I <sub>CC</sub>	IOCD power supply current	From array = high		1.38			1.50	mA
	IOD power supply current	V <sub>IN</sub> = 3V, from array = high		2.53			2.75	mA
ILF	Input load factor					3		Unit load
FO	Fanout							
	To T.S. (I/O only)	(drives 3-state inputs only)				16	16	Inputs
	To Array (I/O only)	(drives internal gates)				10	10	Unit loads
t <sub>pdAV</sub>	Average propagation delay	Fan in = one (1) ISL gate or Schottky buffer		10	14	10	14	ns
	$t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan out = one (1) from 3-state input of an output buffer						
I <sub>CC</sub>	TOC, supply current	V <sub>IN</sub> = 3V, from Array = L		1.76			1.90	mA
	INB, EOCD, supply current	V <sub>IN</sub> = 3V		1.15			1.25	mA
	IOD, supply current	V <sub>IN</sub> = 3V, from Array = H		2.53			2.75	mA
	TEOC, supply current	V <sub>IN</sub> = 3V, from TS = H, from Array = L		3.11			3.35	mA
	TTS, supply current	From Array = L, from TS = H		3.11			3.35	mA
	V <sub>IN</sub> = 3V							
V <sub>TH</sub>	Input threshold voltage	0.80		2.0	0.80		2.0	V
V <sub>CD</sub>	Input clamp diode voltage	I <sub>IN</sub> = -18mA		-1.5			-1.5	V
I <sub>IL</sub>	Input low current	V <sub>IN</sub> = 0.4V		-20			-20	μA
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = 2.7V		20			20	μA
I <sub>I</sub>	Max input high current	V <sub>IN</sub> = 5.5V, V <sub>CC</sub> = Max		100			100	μA
FO	INB & IOD "to array" outputs			10			10	Unit load
	EOCD & IOD "to 3-state" outputs			16			16	Inputs
t <sub>pdLH</sub>	Propagation delay, low-to-high F.O. = one (1) ISL load	See Figure 5a		5	8	5	8	ns
t <sub>pdHL</sub>	Propagation delay, high-to-low F.O. = one (1) ISL load			2	4	2	4	ns
t <sub>pdLH</sub>	Propagation delay, low-to-high F.O. = ten (10) ISL loads			3	4	3	4	ns
t <sub>pdHL</sub>	Propagation delay, high-to-low F.O. = ten (10) ISL loads			4	5	4	5	ns
<b>OUTPUT DRIVER AT V<sub>CC</sub> = 5V</b>								
I <sub>CC</sub>	Power supply current	From array = high		1.38			1.50	mA
ILF	Input load factor	3				3		Unit loads
V <sub>OL</sub>	Output low voltage	I <sub>OL</sub> = 8mA		500				mV
		I <sub>OL</sub> = 4mA				400	mV	
V <sub>OH</sub>	Output high voltage	I <sub>OH</sub> = -400μA	2.7		2.5			V
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0V	-15		-100		-100	mA
t <sub>pdLH</sub>	Propagation delay, low to high output	See Figure 5b		4	8	4	8	ns
t <sub>pdHL</sub>	Propagation delay, high to low output			4	8	4	8	ns

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## ISL GATE ARRAY

8A1200

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
<b>OPEN COLLECTOR OUTPUT BUFFERS: OC, TOC (from array), EOC, TEOC (from array)</b>									
I <sub>CC</sub>	OC power supply current	From array = high			1.38			1.50	mA
	TOC power supply current	From array = low			1.76			1.90	mA
	EOC power supply current	From array = low, from T.S. = high			1.96			2.10	mA
	TEOC power supply current	From array = low, from T.S. = high			3.11			3.35	mA
ILF	Input load factor "from array"		3				3		Unit load
	Input load factor "from T.S."		3				3		Unit load
V <sub>OL</sub>	Output low voltage	I <sub>OL</sub> = 8mA comm			500				mV
		I <sub>OL</sub> = 4mA					400		mV
I <sub>OH</sub>	Output high current	V <sub>OUT</sub> = 5.5V			20			20	μA
t <sub>pdLH</sub>	Propagation delay low to high output	See Figure 5c		9	TBD		9	TBD	ns
t <sub>pdHL</sub>	Propagation delay high to low output			8	TBD		8	TBD	ns
<b>THREE-STATE OUTPUT BUFFERS: TS, TTS (from array)</b>									
I <sub>CC</sub>	TS power supply current	From T.S. = high From array = low			1.96			2.10	mA
	TTS power supply current	From array = low V <sub>IN</sub> = 3V, from T.S. = high			3.11			3.35	mA
ILF	Input load factor, either input		3				3		Unit load
V <sub>OL</sub>	Output low voltage	I <sub>OL</sub> = 8mA			500				mV
		I <sub>OL</sub> = 4mA					400		mV
V <sub>OH</sub>	Output high voltage	I <sub>OH</sub> = -400μA	2.7				2.5		V
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0V	-15		-100		-15		mA
I <sub>OZL</sub>	Three-state off current, output low	V <sub>OUT</sub> = 0.4V			-20			-20	μA
I <sub>OZH</sub>	Three-state off current, output high	V <sub>OUT</sub> = 2.4V			20			20	μA
t <sub>pdLH</sub>	Propagation delay, low to high output	See Figure 5d		4	9		4	9	ns
t <sub>pdHL</sub>	Propagation delay, high to low output			6	10		6	10	ns
t <sub>pdZL</sub>	Propagation delay, HI Z to low output			11	14		11	14	ns
t <sub>pdZH</sub>	Propagation delay, HI Z to high output			10	13		10	13	ns
t <sub>pdLZ</sub>	Propagation delay, low to HI Z output			6	12		6	12	ns
t <sub>pdHZ</sub>	Propagation delay, high to HI Z output			7	7		7	7	ns

## Notes:

- Maximum power dissipation limit of circuit is determined by package selection.
- Guaranteed value is t<sub>pdAV</sub>.
- For all input parameters on TEOC and TTS, the "from Three-State" input should be high.

**8A1200 EVALUATION CIRCUIT****CG1001****PRODUCT DESCRIPTION**

The 8A1200/CG1001 Evaluation Circuit is a committed array of ISL gates, Schottky buffers, and LSTTL I/O cells, providing the user with several logic functions that can be easily and economically implemented by the use of semi-custom LSI. Basically, the CG1001 provides a demonstration vehicle for characterizing design functions of the 8A1200 ISL Gate Array; the demonstration part contains logic functions that are representative of, and can be compared with, those of standard 7400-series parts.

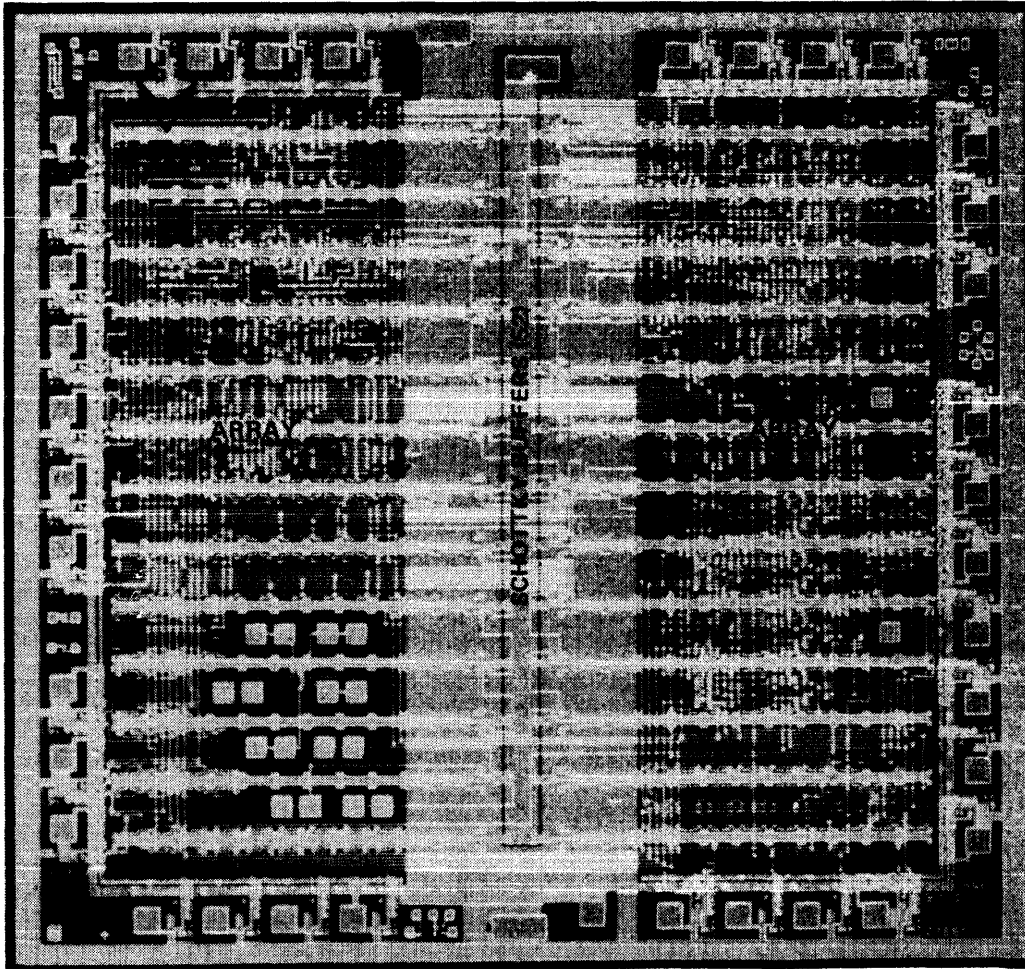
Also, the CG1001 contains several test configurations that can be used in evaluating circuit performance under various logical, topological, and environmental conditions. A block diagram of the Gate Array Evaluation Circuit is shown in Figure 2 and logic representations of each discrete function are shown in Figures 3 through 6.

**FEATURES****REPRESENTATIVE LOGIC FUNCTIONS:**

- 8 to 1 Multiplexer (Figure 3)—similar to 74152.
- 4-Bit Adder (Figure 4)—similar to 7483.
- 4-Bit Universal Shift Register (Figure 5)—similar to 74194

**SPECIAL TEST CIRCUITS**

- D-flip flop wired as a toggle flip flop
- Demonstration of fanout effects on ISL gates
- Test of fanin and pattern sensitivity effects on ISL gates
- Ring oscillators which show the basic gate delays of ISL gates and Schottky buffers under various layout and logical conditions

**8A1200/CG1001 EVALUATION CIRCUIT**

8A1200 EVALUATION CIRCUIT

CG1001

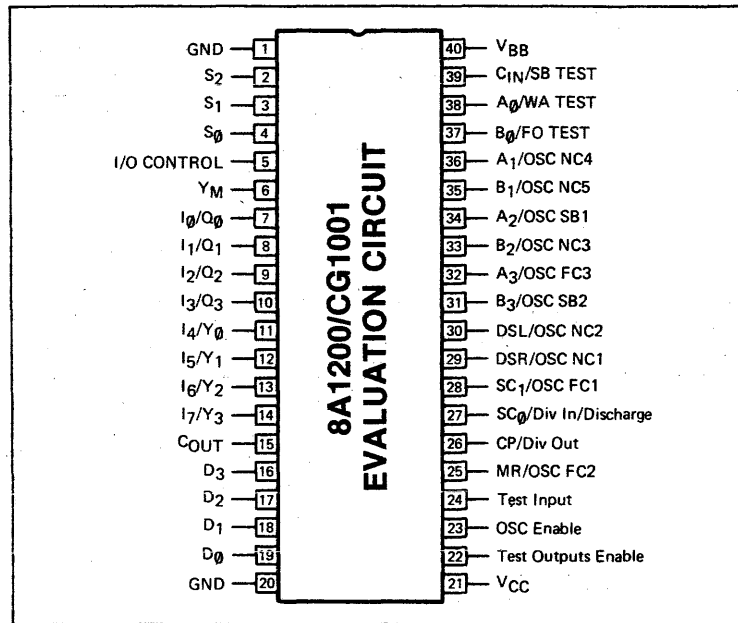


Figure 1. Package Configuration and Pin Designators of 8A1200/CG1001 Evaluation Circuit

DC ELECTRICAL CHARACTERISTICS

PARAMETER	DESCRIPTION	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
$I_{CC}$	Supply current at $V_{CC}$	Pins 5 and 22 = L (OUTPUT ENABLES)		56	87		59	93	mA
$I_{BB}$	Supply current at $V_{BB}$			62	85		62	93	mA

Note. All other DC CHARACTERISTICS are specific to the I/O cells and can be found in the Data Sheet pertaining to the 8A1200 ISL Gate Array.

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**8A1200 EVALUATION CIRCUIT**

**CG1001**

**AC ELECTRICAL CHARACTERISTICS**

PARAMETERS (Note 1)	REFERENCES		TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
	FROM	TO		MIN	TYP	MAX	MIN	TYP	MAX	
Propagation delay of 8-to-1 Multiplexer:										
$t_{pdHL}$	PINS 7-14	PIN 6	See test setup		40					ns
$t_{pdLH}$	PINS 7-14	PIN 6	below		55					ns

**TEST SETUP FOR 8-TO-1 MULTIPLEXER (TEST OUTPUTS OFF)**

FUNCTION →	S <sub>2</sub>	S <sub>1</sub>	S <sub>0</sub>	I <sub>0</sub>	I <sub>1</sub>	I <sub>2</sub>	I <sub>3</sub>	I <sub>4</sub>	I <sub>5</sub>	I <sub>6</sub>	I <sub>7</sub>	Y <sub>M</sub>
PIN NO. →	2	3	4	7	8	9	10	11	12	13	14	6
STATES →	L	L	L	In	L	L	L	L	L	L	L	Out
	L	L	H	L	In	L	L	L	L	L	L	Out
	L	H	L	L	L	In	L	L	L	L	L	Out
	L	H	H	L	L	L	In	L	L	L	L	Out
	H	L	L	L	L	L	L	In	L	L	L	Out
	H	L	H	L	L	L	L	L	In	L	L	Out
	H	H	L	L	L	L	L	L	L	In	L	Out
	H	H	H	L	L	L	L	L	L	L	In	Out

PARAMETERS (Note 1)	REFERENCES		TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
	FROM	TO		MIN	TYP	MAX	MIN	TYP	MAX	
Propagation delay of 4-Bit Adder:										
$t_{pdHL}$	PIN 39	PIN 15	See test setup below		41					ns
$t_{pdLH}$				40		ns				
$t_{pdHL}$	PIN 39	PIN 11		42		ns				
$t_{pdLH}$				35		ns				
$t_{pdHL}$	PIN 32	PIN 15		40		ns				
$t_{pdLH}$				35		ns				

**TEST SETUP FOR 4-BIT ADDER (TEST OUTPUTS OFF)**

FUNCTION →	B <sub>3</sub>	A <sub>3</sub>	B <sub>2</sub>	A <sub>2</sub>	B <sub>1</sub>	A <sub>1</sub>	B <sub>0</sub>	A <sub>0</sub>	C <sub>IN</sub>	Y <sub>3</sub>	Y <sub>2</sub>	Y <sub>1</sub>	Y <sub>0</sub>	C <sub>OUT</sub>
PIN NO. →	31	32	33	34	35	36	37	38	39	14	13	12	11	15
STATES →	L	H	L	H	L	H	L	H	In	X	X	X	X	Out
	L	L	L	L	L	L	L	L	In	X	X	X	Out	X
	H	In	L	H	L	H	L	H	L	X	X	X	X	Out

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**8A1200 EVALUATION CIRCUIT**

**CG1001**

**AC ELECTRICAL CHARACTERISTICS (cont'd)**

PARAMETERS (Note 1)	REFERENCES		TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
	FROM	TO		MIN	TYP	MAX	MIN	TYP	MAX	
Propagation delay of Fan-In/Fan-Out tests:										
$t_{pdHL}$	PIN 24	PIN 37	See test setup		37					ns
$t_{pdLH}$			"r" below		36					ns
$t_{pdHL}$			See test setups		38					ns
$t_{pdLH}$			"s" and "t" below		35					ns
$t_{pdHL}$			See test setups		37					ns
$t_{pdLH}$			"v" and "w" below		33					ns
$t_{pdHL}$	PIN 24	PIN 38	See test setup		744					ns
$t_{pdLH}$			"p" below		737					ns
$t_{pdHL}$			See test setup		39					ns
$t_{pdLH}$			"r" below		40					ns
$t_{pdHL}$			See test setups		40					ns
$t_{pdLH}$			"s", "t", and "u" below		40					ns
$t_{pdHL}$			See test setups		51					ns
$t_{pdLH}$			"v", "w", and "x" below		40					ns
$t_{pdHL}$	PIN 24	PIN 39	See test setup		35					ns
$t_{pdLH}$			"r" below		35					ns
$t_{pdHL}$			See test setup		34					ns
$t_{pdLH}$			"s" below		36					ns
$t_{pdHL}$			See test setup		28					ns
$t_{pdLH}$			"v" below		38					ns

**TEST SETUPS FOR FAN-IN/FAN-OUT TESTS (See DELAY COMPARISONS)**

TEST SETUP	SELECTOR		DISCHARGE (NOTE 2) PIN 27	LOAD PIN 2	TEST INPUT PIN 24	FAN-OUT TEST PIN 37	Wired-AND TEST PIN 38	SCHOTTKY BUFFER TEST PIN 39
	S <sub>1</sub> PIN 3	S <sub>0</sub> PIN 4						
p	L	L	L	L	In	-	Out	-
r	L	H	L	L	In	Out	Out	Out
s	H	L	L	L	In	Out	Out	Out
t	H	L		H	In	Out	Out	-
u	H	L		L	In	-	Out	-
v	H	H	L	L	In	Out	Out	Out
w	H	H		H	In	Out	Out	-
x	H	H		L	In	-	Out	-

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**8A1200 EVALUATION CIRCUIT**

**CG1001**

**AC ELECTRICAL CHARACTERISTICS (cont'd)**

PARAMETERS (Note 1)	REFERENCES		TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
	FROM	TO		MIN	TYP	MAX	MIN	TYP	MAX	
Propagation delay of 4-Bit Shift Register:			See test setup below							
$t_{pdHL}$	PIN 26	PINS 7,8			34					ns
$t_{pdLH}$					34					ns
$t_{pdHL}$	PIN 26	PIN 9			37					ns
$t_{pdLH}$					34					ns
$t_{pdHL}$	PIN 26	PIN 10			33					ns
$t_{pdLH}$					33					ns
$t_{pdHL}$	PIN 25	PINS 7-10			33					ns

FUNCTION (NOTES 3, 4, & 5) →	CP	$\overline{MR}$	$SC_0$	$SC_1$	$D_{ST}$	$D_{SL}$	$D_0$	$D_1$	$D_2$	$D_3$	$Q_0$	$Q_1$	$Q_2$	$Q_3$
PIN NO. →	26	25	27	28	29	30	19	18	17	16	7	8	9	10
STATES } →	In(↑)	H	H	H	X	X	$D_{IN}$	X	X	X	Out	X	X	X
	In(↑)	H	H	H	X	X	X	X	$D_{IN}$	X	X	X	Out	X
	In(↑)	H	H	H	X	X	X	X	X	$D_{IN}$	X	X	X	Out
	X	In	X	X	X	X	X	X	X	X	Out	Out	Out	Out

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# 8A1200 EVALUATION CIRCUIT

CG1001

## AC ELECTRICAL CHARACTERISTICS (cont'd)

PARAMETERS (Note 1)	REFERENCES		TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS
	FROM	TO		MIN	TYP	MAX	MIN	TYP	MAX	
Propagation delay for Divider:										
$t_{pdHL}$	PIN 27	PIN 26	See test setup below		44					ns
$t_{pdLH}$					39					ns
Period of Ring Oscillator:										
tosc	—	PIN 25	See test setup below		108					ns
tosc	—	PIN 28			109					ns
tosc	—	PIN 29			107					ns
tosc	—	PIN 30			106					ns
tosc	—	PIN 31			101					ns
tosc	—	PIN 32			107					ns
tosc	—	PIN 33			86					ns
tosc	—	PIN 34			91					ns
tosc	—	PIN 35			147					ns
tosc	—	PIN 36			82					ns

**TEST SETUP FOR DIVIDER/RING OSCILLATORS**

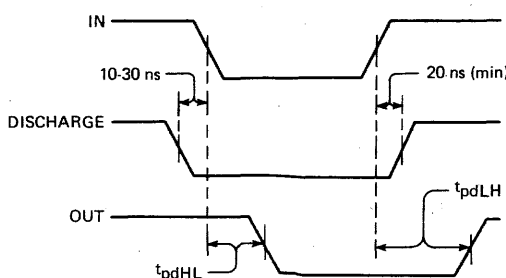
PIN 8 = L (INPUT)  
 PIN 22 = H (TEST OUTPUTS ON)  
 PIN 23 = L (OSC OFF)  
 PIN 23 = H (OSC ON)

DIV IN PIN 27	DIV OUT PIN 26	OSC FC2 PIN 25	OSC FC1 PIN 28	OSC NC1 PIN 29	OSC NC2 PIN 30	OSC SB2 PIN 31	OSC FC3 PIN 32	OSC NC3 PIN 33	OSC SB1 PIN 34	OSC NC5 PIN 35	OSC NC4 PIN 36
In(t)	Out	Oscillator Calculations: The ten 11-gate rings oscillate with a period ( $t_{osc}$ ) equal to 22 gate delays. Average gate delay ( $t_{avg}$ = half pair delay) can be calculated as follows: $t_{avg} = \frac{t_{osc}}{22} = \frac{t_{pdLH} + t_{pdHL}}{2}$									

Notes:

1. Measure  $t_{pdLH}$  and  $t_{pdHL}$  from "In" to "Out" for each path.
2. Discharge input (pin 27) must meet both MIN and MAX times for setup and hold— see WAVEFORM 1.
3. For  $Q_0$  and  $Q_1$  outputs (pins 7 and 8), propagation delay is representative of the delay through an input buffer, a three-state output buffer with a fan-in of 1, and a standard ISL "D" flip-flop.
4. For the  $Q_2$  output (pin 9), the propagation delay will differ from that of  $Q_0$  and  $Q_1$  by the  $\Delta$  time delay caused by the additional fan-in of 4 on the three-state output buffer.
5. For the  $Q_3$  output (pin 10), the propagation delay will differ from that of  $Q_0$  and  $Q_1$  by the  $\Delta$  time delay caused by the additional fan-in of 2 on the three-state output buffer.

### WAVEFORM 1: Discharge Input Timing



### DELAY COMPARISONS

Wired-AND Test (Pin 38)	
COMPARISON (Note 1)	DESCRIPTION
(r) - (s)	Effect of $\Delta$ fan-in of 4 on ISL gate delay, load capacitors precharged
(r) - (u)	Effect of $\Delta$ fan-in of 4 on ISL gate delay, load capacitors discharged
(s) - (v)	Effect of $\Delta$ fan-in of 5 on ISL gate delay, load capacitors precharged
(u) - (x)	Effect of $\Delta$ fan-in of 5 on ISL gate delay, load capacitors discharged
(r) - (v)	Effect of $\Delta$ fan-in of 9 on ISL gate delay, load capacitors precharged
(r) - (x)	Effect of $\Delta$ fan-in of 9 on ISL gate delay, load capacitors discharged
(t) - (u)	Effect of dummy loads, fan-in = 5
(w) - (x)	Effect of dummy loads, fan-in = 10
(s) - (u)	Effect of worst case pattern sensitivity, fan-in = 5
(v) - (x)	Effect of worst case pattern sensitivity, fan-in = 10
(p)	Delay of 142 ISL gates + input buffer + T.S. output buffer

**ISL GATE ARRAY**

**8A1260**

**PRODUCT DESCRIPTION**

The 8A1260 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2), Schottky buffers (Figure 3) and LSTTL-compatible I/O cells (Figure 4). Thus, up to 1200 gates can be custom interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A1260 array is based on a technological subset of LSI called ISL (Integrated Schottky Logic). ISL combines the features of Schottky and the density of I<sup>2</sup>L Bipolar technologies.

Designing with the 8A1260 is easy and fast, requiring no more than conventional logic design, logic simulation, and custom coding of metal interconnections among preprocessed logic gates on the array. Refer to Table 1 for a comparison of ISL and 74LS logic functions. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting 1144 ISL NAND gates, using two layers of metal routing. Fifty-two Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 60 LSTTL I/O buffers can be specified. Each I/O can

be configured as 1-of-4 input buffers, 1-of-4 output buffers, or as a combination of one input buffer and one output buffer for a transceiver.

**FEATURES**

- Customer programmable LSI
- 1144 ISL (NAND) gates
- Two-layer metal interconnection
- 52 Schottky buffers
- 60 I/O buffers
- LSTTL compatible
- Standard PNP inputs
- 8mA output current sink
- -55°C to +125°C ambient temperature
- 4ns gate speed (typical)
- Speed-power product—0.7 picojoules
- 68 pin package

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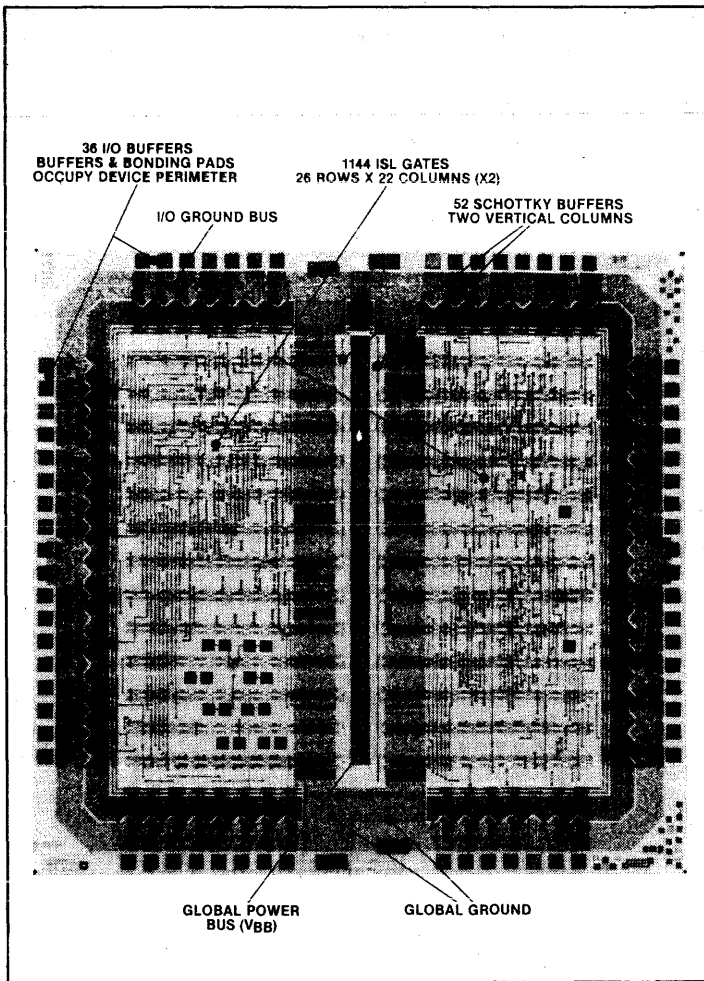


Figure 1. Internal Configuration of 8A1260

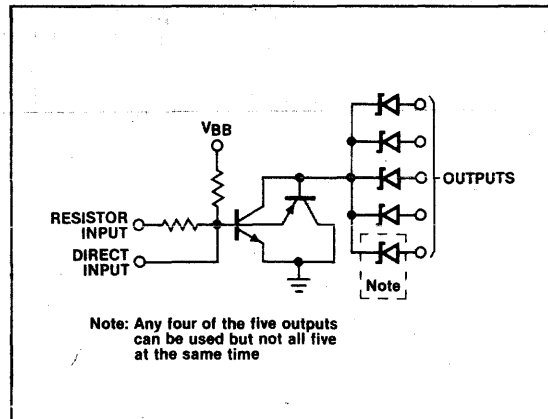


Figure 2. ISL Gate—Schematic Diagram

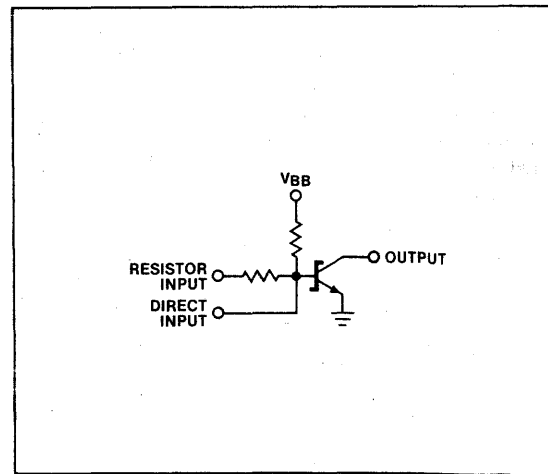


Figure 3. Schottky Buffer—Schematic Diagram

**ISL GATE ARRAY**

**8A1260**

**AC AND DC ELECTRICAL CHARACTERISTICS**

**Conditions:**

Commercial—	Military—
V <sub>CC</sub> = 5.0V (± 5%)	V <sub>CC</sub> = 5.0V (± 10%)
V <sub>BB</sub> = 1.5V (± 10%)	V <sub>BB</sub> = 1.5V (± 10%)
T <sub>A</sub> = 0°C to 70°C	T <sub>A</sub> = -55°C to 125°C

**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	DESCRIPTION	RATING	UNIT	PARAMETER	DESCRIPTION	RATING	UNIT
V <sub>CC</sub>	Supply voltage	+7.0	V	V <sub>O</sub>	Voltage applied to open-collector output in off-state	-0.5 to +7.0	V
V <sub>BB</sub>	ISL gate supply voltage	+7.0	V	T <sub>A</sub>	Ambient temperature, operating	-55 to +125	°C
E <sub>IN</sub>	Input voltage, continuous	-0.5 to +5.5	V	T <sub>STG</sub>	Storage temperature	-65 to +150	°C
I <sub>iN</sub>	Input current, continuous	-30 to +1.0	mA				

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN.	TYP	MAX	MIN	TYP	MAX		
<b>ISL GATE ARRAY</b>									
I <sub>BB/G</sub>	Power supply current per gate			190			190	μA	
ILF	Input load factor			1			1	Unit load	
FO	Fan-out			4			4	Unit load	
t <sub>pdAV</sub>	Average gate propagation delay $t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = one (1) ISL gate or Schottky buffer Fan-out = one (1) ISL gate or Schottky buffer		4	6		4	6	ns
t <sub>pdHL</sub>	High-to-low propagation delay	Delay is inferred from circuit simulation		1	2		1	2	ns
t <sub>pdLH</sub>	Low-to-high propagation delay			7	10		7	10	ns

ISL GATE ARRAY

8A1260

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
$I_{BB/G}$	Power supply current per gate			190			190	$\mu A$	
ILF	Input load factor			1			1	Unit load	
FO	Fan-out			10			10	Unit load	
$t_{pdAV}$	Average gate propagation delay $t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = one (1) ISL gate or Schottky buffer Fan-out = one (1) ISL gate or Schottky buffer		4	6		4	6	ns
$t_{pdHL}$	High-to-low propagation delay	Delay is inferred from circuit simulation		1	2		1	2	ns
$t_{pdLH}$	Low-to-high propagation delay			7	10		7	10	ns
$I_{CC}$	Power supply current	$V_{IN} = 3V (IN \text{ from Array} = H)$			1.30		1.40	mA	
$V_{TH}$	Input threshold voltage		0.80		2.0	0.80	2.0	V	
VCD	Input clamp diode voltage	$I_{IN} = -18mA$			-1.5		-1.5	V	
$I_{IL}$	Input low current	$V_{IN} = 0.4V$			-20		-20	$\mu A$	
$I_{IH}$	Input high current	$V_{IN} = 2.7V$			20		20	$\mu A$	
$I_I$	Max input high current	$V_{IN} = 5.5V, V_{CC} = \text{Max}$			100		100	$\mu A$	
FO	INB & IOD "to array" outputs				10		10	Unit load	
	EOCD & IOD "to three-state" outputs				16		16	Unit load	
$t_{pdLH}$	Propagation delay, low-to-high F.O. = one (1) ISL load	(See Fig. 5a)		5	8		5	8	ns
$t_{pdHL}$	Propagation delay, high-to-low F.O. = one (1) ISL load			2	4		2	4	ns
$t_{pdLH}$	Propagation delay, low-to-high F.O. = ten (10) ISL loads			3	4		3	4	ns
$t_{pdHL}$	Propagation delay, high-to-low F.O. = ten (10) ISL loads			4	5		4	5	ns
<b>OUTPUT BUFFER, AP (Active Pullup)</b>									
$I_{CC}$	Power supply current	From array = high			1.38		1.50	mA	
ILF	Input load factor		3			3		Unit loads	
$V_{OL}$	Output low voltage	$I_{OL} = 8mA$ $I_{OL} = 4mA$			500		400	mV mV	
$V_{OH}$	Output high voltage	$I_{OH} = -400\mu A$	2.7			2.5		V	
$I_{OS}$	Output short circuit current	$V_{OUT} = 0V$	-15		-100	-15	-100	mA	

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**ISL GATE ARRAY**

**8A1260**

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
<b>OUTPUT BUFFER: AP (Active Pullup) (continued)</b>									
t <sub>pdLH</sub>	Propagation delay, low-to-high output	(See Fig. 5b)	4	8		4	8	ns	
t <sub>pdHL</sub>	Propagation delay, high-to-low output		4	8		4	8	ns	
<b>OPEN COLLECTOR OUTPUT BUFFER</b>									
I <sub>CC</sub>	OC power supply current EOC power supply current	From array = high From array = low, From T.S. = high		1.38 1.98			1.50 2.10	mA mA	
ILF	Input load factor "from array" Input load factor "from T.S."		3 3		3 3			Unit load Unit load	
VOL	Output low voltage	I <sub>OL</sub> = 8mA I <sub>OL</sub> = 4mA		500			400	mV mV	
IOH	Output high current	V <sub>OUT</sub> = 5.5V		20			20	μA	
t <sub>pdLH</sub>	Propagation delay, low-to-high output	(See Fig. 5c)	9	TBD		9	TBD	ns	
t <sub>pdHL</sub>	Propagation delay, high-to-low output		8	TBD		8	TBD	ns	
<b>THREE-STATE OUTPUT BUFFER</b>									
I <sub>CC</sub>	TS power supply current	From T.S. = high From array = low		1.98			2.10	mA	
ILF	Input load factor, either input		3		3			Unit load	
VOL	Output low voltage	I <sub>OL</sub> = 8mA I <sub>OL</sub> = 4mA		500			400	mV mV	
VOH	Output high voltage	I <sub>OH</sub> = -400μA	2.7		2.5			V	
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0V	-15	-100	-15		-100	mA	
I <sub>OLZ</sub>	Three-state off current, output low	V <sub>OUT</sub> = 0.4V		-20			-20	μA	
I <sub>OHZ</sub>	Three-state off current,	V <sub>OUT</sub> = 2.4V		20			20	μA	
t <sub>pdLH</sub>	Propagation delay, low-to-high output (Note)	R <sub>L</sub> = 2K		4	9		4	9	ns
t <sub>pdHL</sub>	Propagation delay, high-to-low output (Note)	C <sub>L</sub> = 15pf		6	10		6	10	ns
t <sub>pdZL</sub>	Propagation delay, HI-Z to low output	(See Fig. 5d)		11	14		11	14	ns
t <sub>pdZH</sub>	Propagation delay, HI-Z to high output			10	13		10	13	ns
t <sub>pdLZ</sub>	Propagation delay, LOW to HI-Z output			6	12		6	12	ns
t <sub>pdHZ</sub>	Propagation delay, high to HI-Z output			7	7		7	7	ns

Note: Guaranteed value is t<sub>pd(Ave)</sub>

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**ISL GATE ARRAY**

**8A1542**

**PRODUCT DESCRIPTION**

The 8A1542 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2), Schottky buffers (Figure 3) and LSTTL-compatible I/O cells (Figure 4). Thus, up to 1400 gates can be custom interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A1542 array is based on a technological subset of LSI called ISL (Integrated Schottky Logic). ISL combines the best features of low-power Schottky and I<sup>2</sup>L Bipolar technologies.

Designing with the 8A1542 is easy and fast, requiring no more than conventional logic design, logic simulation, and custom coding of metal interconnections among preprocessed logic gates on the array—refer to Table 1 for a comparison of ISL and 74LS logic functions. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting 1408 ISL NAND gates, using two layers of metal routing. Sixty-four Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 42 LSTTL I/O buffers can be specified. As shown in Figure 4, each I/O can be configured to implement any one of 11 different functions: inputs, input/output, totem-pole, open collector, and three-state.

**FEATURES**

- Customer programmable LSI
- 1408 ISL (NAND) gates
- Two-layer metal interconnection
- 64 Schottky buffers
- 42 I/O buffers
- LSTTL compatible
- Standard PNP inputs
- 8mA output current sink
- -55°C to +125°C ambient temperature
- 4ns gate speed (typical)
- Speed-power product—0.7 picojoules
- 28, 40, or 44-pin package

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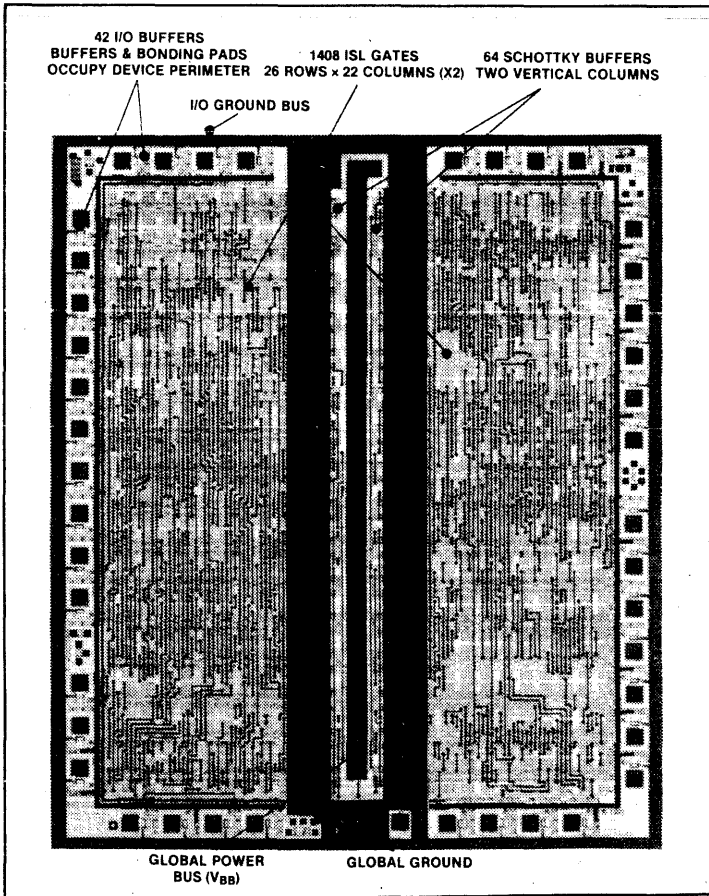


Figure 1. Internal Configuration of 8A1542 ISL Gate Array

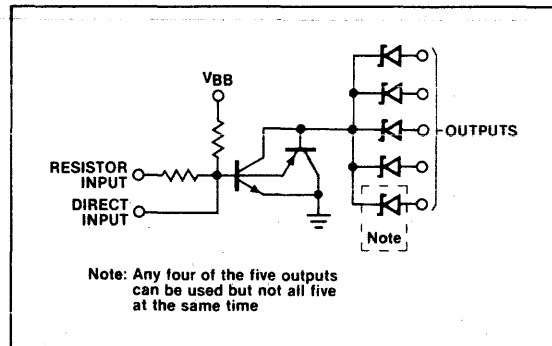


Figure 2. ISL Gate—Schematic Diagram

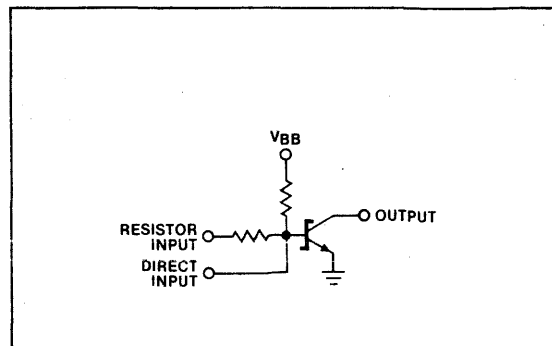


Figure 3. Schottky Buffer—Schematic Diagram

**ISL GATE ARRAY**

**8A1542**

**AC AND DC ELECTRICAL CHARACTERISTICS**

Conditions: Commercial— Military—  
 $V_{CC} = 5.0V (\pm 5\%)$   $V_{CC} = 5.0V (\pm 10\%)$   
 $V_{BB} = 1.5V (\pm 10\%)$   $V_{BB} = 1.5V (\pm 10\%)$   
 $T_A^1 = 0^\circ C \text{ to } 70^\circ C$   $T_A^1 = -55^\circ C \text{ to } 125^\circ C$

**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	DESCRIPTION	RATING	UNIT	PARAMETER	DESCRIPTION	RATING	UNIT
V <sub>CC</sub>	Supply voltage	+7.0	V	V <sub>O</sub>	Voltage applied to open-collector output in off-state	-0.5 to +7.0	V
V <sub>BB</sub>	ISL gate supply voltage	+7.0	V	T <sub>A</sub>	Ambient temperature, operating	-55 to +125	°C
E <sub>IN</sub>	Input voltage, continuous	-0.5 to +5.5	V	T <sub>STG</sub>	Storage temperature	-65 to +150	°C
I <sub>IN</sub>	Input current, continuous	-30 to +1.0	mA				

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
<b>ISL GATE ARRAY</b>									
I <sub>BB/G</sub>	Power supply current per gate			190			190	μA	
ILF	Input load factor			1			1	Unit load	
FO	Fanout			4			4	Unit load	
t <sub>pdAV</sub>	Average gate propagation delay $t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = one (1) ISL gate or Schottky buffer Fan-out = one (1) ISL gate or Schottky buffer		4	6		4	6	ns
t <sub>pdHL2</sub>	High-to-low propagation delay	Delay is inferred from circuit simulation		1	2		1	2	ns
t <sub>pdLH2</sub>	Low-to-high propagation delay			7	10		7	10	ns
<b>SCHOTTKY BUFFER (internal)</b>									
I <sub>BB/G</sub>	Power supply current per gate			190			190	μA	
ILF	Input load factor			1			1	Unit load	
FO	Fanout			10			10	Unit load	
t <sub>pdAV</sub>	Average gate propagation delay $t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = one (1) ISL gate or Schottky buffer Fan-out = one (1) ISL gate or Schottky buffer		4	6		4	6	ns
t <sub>pdHL2</sub>	High-to-low propagation delay	Delay is inferred from circuit simulation		1	2		1	2	ns
t <sub>pdLH2</sub>	Low-to-high propagation delay			7	10		7	10	ns

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**ISL GATE ARRAY**

**8A1542**

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
I <sub>CC</sub>	IOCD power supply current	From array = high			1.38				mA
	IOD power supply current	V <sub>IN</sub> = 3V, from array = high			2.53				mA
ILF	Input load factor	3			3			Unit load	
FO	Fanout								
	To T.S. (I/O only)	(drives 3-state inputs only)			16	16			Inputs
	To Array (I/O only)	(drives internal gates)			10	10			Unit loads
t <sub>pdAV</sub>	Average propagation delay	Fan in = one (1) ISL gate or Schottky buffer			10	14	10	14	ns
	$t_{pdAV} = \frac{t_{pdLH} + t_{pdHL}}{2}$	Fan out = one (1) from 3-state input of an output buffer							
I <sub>CC</sub>	TOC, supply current	V <sub>IN</sub> = 3V, from Array = L			1.76	1.90			mA
	INB, EOCD, supply current	V <sub>IN</sub> = 3V			1.15	1.25			mA
	IOD, supply current	V <sub>IN</sub> = 3V, from Array = H			2.53	2.75			mA
	TEOC, supply current	V <sub>IN</sub> = 3V, from TS = H, from Array = L			3.11	3.35			mA
	TTS, supply current	From Array = L, from TS = H V <sub>IN</sub> = 3V			3.11	3.35			mA
V <sub>TH</sub>	Input threshold voltage	0.8			2.0	0.80	2.0		V
V <sub>CD</sub>	Input clamp diode voltage	I <sub>IN</sub> = -18mA			-1.5	-1.5			V
I <sub>IL</sub>	Input low current	V <sub>IN</sub> = 0.4V			-20	-20			μA
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = 2.7V			20	20			μA
I <sub>I</sub>	Max input high current	V <sub>IN</sub> = 5.5V, V <sub>CC</sub> = Max			100	100			μA
FO	INB & IOD "to array" outputs				10	10			Unit load
	EOCD & IOD "to 3-state" outputs				16	16			Inputs
t <sub>pdLH</sub>	Propagation delay, low-to-high F.O. = one (1) ISL load				5	8	5	8	ns
t <sub>pdHL</sub>	Propagation delay, high-to-low F.O. = one (1) ISL load				2	4	2	4	ns
t <sub>pdLH</sub>	Propagation delay, low-to-high F.O. = ten (10) ISL loads				3	4	3	4	ns
t <sub>pdHL</sub>	Propagation delay, high-to-low F.O. = ten (10) ISL loads				4	5	4	5	ns
<b>OUTPUT BUFFER - AS ACTIVE DRIVER</b>									
I <sub>CC</sub>	Power supply current	From array = high			1.38	1.50			mA
ILF	Input load factor	3			3			Unit loads	
V <sub>OL</sub>	Output low voltage	I <sub>OL</sub> = 8mA			500				mV
		I <sub>OL</sub> = 4mA				400			mV
V <sub>OH</sub>	Output high voltage	I <sub>OH</sub> = -400μA			2.7	2.5			V
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0V			-15	-100	-15	-100	mA
t <sub>pdLH</sub>	Propagation delay, low to high output				4	8	4	8	ns
t <sub>pdHL</sub>	Propagation delay, high to low output				4	8	4	8	ns

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# ISL GATE ARRAY

## 8A1542

PARAMETER	TEST CONDITIONS	LIMITS (COMMERCIAL)			LIMITS (MILITARY)			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX		
<b>OPEN COLLECTOR OUTPUT BUFFERS: OC, TOC (from array), EOC, TEOC (from array)</b>								Preliminary	
I <sub>CC</sub>	OC power supply current	From array = high			1.38			1.50	mA
	TOC power supply current	From array = low			1.76			1.90	mA
	EOC power supply current	From array = low, from T.S. = high			1.96			2.10	mA
	TEOC power supply current	From array = low, from T.S. = high			3.11			3.35	mA
ILF	Input load factor "from array"		3			3			Unit load
	Input load factor "from T.S."		3			3			Unit load
V <sub>OL</sub>	Output low voltage	I <sub>OL</sub> = 8mA			500				mV
		I <sub>OL</sub> = 4mA					400		mV
I <sub>OH</sub>	Output high current	V <sub>OUT</sub> = 5.5V			20		20		μA
t <sub>pdLH</sub>	Propagation delay low to high output			9	TBD		9	TBD	ns
t <sub>pdHL</sub>	Propagation delay high to low output			8	TBD		8	TBD	ns
<b>THREE-STATE OUTPUT BUFFERS: TS, TTS (from array)</b>									
I <sub>CC</sub>	TS power supply current	From T.S. = high From array = low			1.96			2.10	mA
	TTS power supply current	From array = low V <sub>IN</sub> = 3V, from T.S. = high			3.11			3.35	mA
ILF	Input load factor, either input		3			3			Unit load
V <sub>OL</sub>	Output low voltage	I <sub>OL</sub> = 8mA			500				mV
		I <sub>OL</sub> = 4mA					400		mV
V <sub>OH</sub>	Output high voltage	I <sub>OH</sub> = -400μA	2.7			2.5			V
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0V	-15		-100	-15		-100	mA
I <sub>OZL</sub>	Three-state off current, output low	V <sub>OUT</sub> = 0.4V			-20			-20	μA
I <sub>OZH</sub>	Three-state off current, output high	V <sub>OUT</sub> = 2.4V			20			20	μA
t <sub>pdLH</sub>	Propagation delay, low to high output			4	9		4	9	ns
t <sub>pdHL</sub>	Propagation delay, high to low output			6	10		6	10	ns
t <sub>pdZL</sub>	Propagation delay, HI Z to low output			11	14		11	14	ns
t <sub>pdZH</sub>	Propagation delay, HI Z to high output			10	13		10	13	ns
t <sub>pdLZ</sub>	Propagation delay, low to HI Z output			6	12		6	12	ns
t <sub>pdHZ</sub>	Propagation delay, high to HI Z output			7	7		7	7	ns

**Notes:**

- Maximum power dissipation limit of circuit is determined by package selection.
- Guaranteed value is t<sub>pdAV</sub>.
- For all input parameters on TEOC and TTS, the "from Three-State" input should be high.

**ISL GATE ARRAY**

**8A1664**

**PRODUCT DESCRIPTION**

The 8A1664 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2), Schottky Buffers (Figure 3), and the LSTTL compatible I/O cells. Thus, up to 1600 gates can be custom interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A1664 array is based on a technological subset of LSI called ISL (Integrated Schottky Logic). ISL combines the features of Schottky and the density of I<sup>2</sup>L Bipolar technologies.

Designing with the 8A1664 is easy and fast, requiring no more than conventional logic design, logic simulation, and custom coding of metal interconnections among preprocessed logic gates on the array. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting 1560 ISL NAND gates, using two layers of metal routing. Sixty Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 64 LS TTL I/O buffers can be specified. Each

8-milliampere I/O site can be configured as 1-of-6 input/internal buffers or as 1-of-8 output buffers; each 24-milliampere I/O site can also be configured as 1-of-6 input/internal buffers but the output buffer configuration can be 1-of-12. For a transceiver, either I/O site can be connected in combinations of one input and one output buffer.

**FEATURES**

- Customer programmable LSI
- 1560 ISL (NAND) gates
- Two-layer metal interconnection
- 60 Schottky buffers
- 64 I/O buffers
- LS TTL compatible
- Standard PNP inputs
- 8-, or 24-milliampere output current sink
- -55°C to +125°C ambient temperature
- 4-nanosecond gate speed (typical)
- Speed-power product — 0.7 picojoules
- 68 pin package

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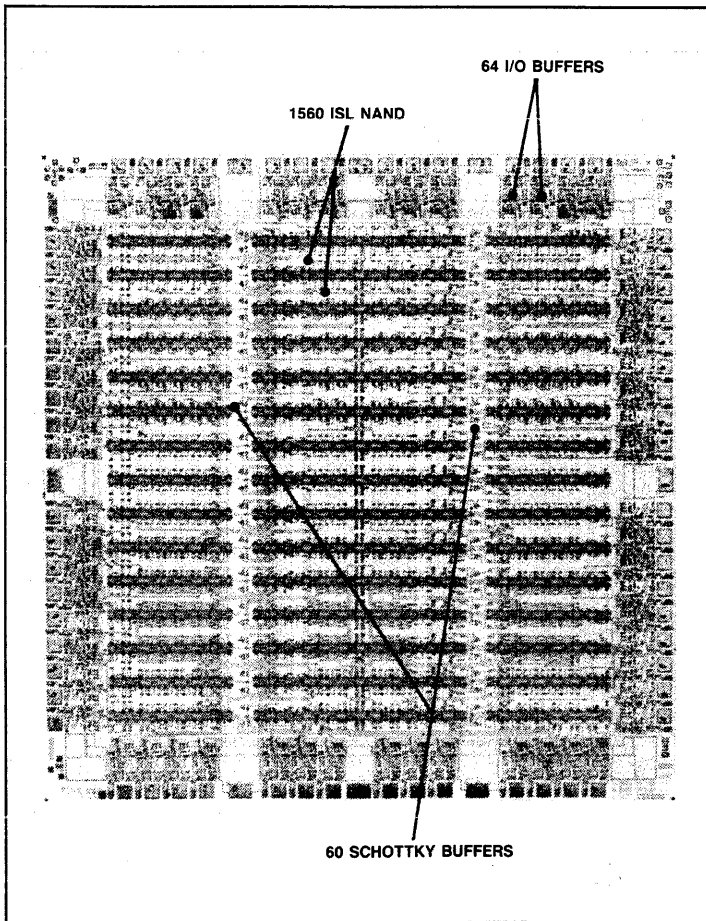


Figure 1. Internal Configuration of 8A1664

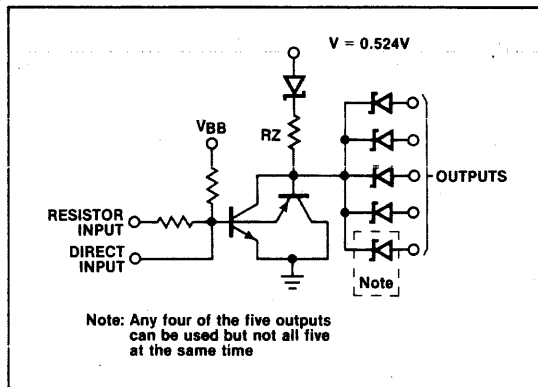


Figure 2. ISL Gate — Schematic Diagram

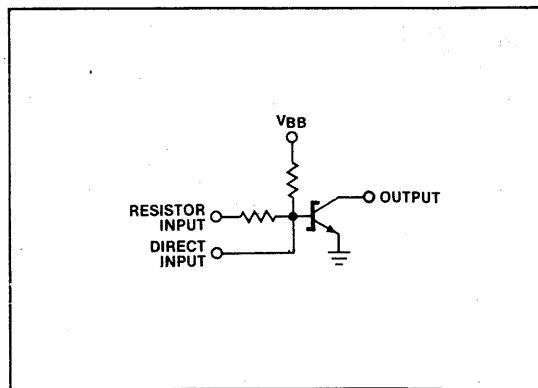


Figure 3. Schottky Buffer — Schematic Diagram

ISL GATE ARRAY

8A1864

PRODUCT DESCRIPTION

The 8A1864 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2). Schottky buffers (Figure 3) and LSTTL-compatible I/O cells. Thus, up to 1740 gates and 60 buffers can be interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A1864 array is based on a technological subset of LSI called ISL (integrated Schottky Logic). ISL combines the features of Schottky and the density of I<sup>2</sup>L Bipolar technologies.

Designing with the 8A1864 is easy and fast, requiring no more than conventional logic design, logic simulation, and coding of metal interconnections between preprocessed logic gates on the array. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting up to 1740 ISL NAND gates, and up to 60 buffers, using two layers of metal routing. Sixty Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 64 LS TTL I/O buffers can be specified. Each 8-milliampere I/O site can be

configured as 1-of-6 input/internal buffers or as 1-of-8 output buffers; each 24-milliampere I/O site can also be configured as 1-of-6 input/internal buffers but the output buffer configuration can be 1-of-12. For a transceiver, either I/O site can be connected in combinations of one input and one output buffer.

FEATURES

- Customer programmable LSI
- 1740 ISL (NAND) gates
- Two-layer metal interconnection
- 60 Schottky buffers
- 72 I/O buffers
- LS TTL compatible
- Standard PNP inputs
- 8-, or 24-milliampere output current sink
- -55°C to +125°C ambient temperature
- 4-nanosecond gate speed (typical)
- Speed-power product — 0.7 picojoules
- 40-, 44-, 50- or 68-pin packages

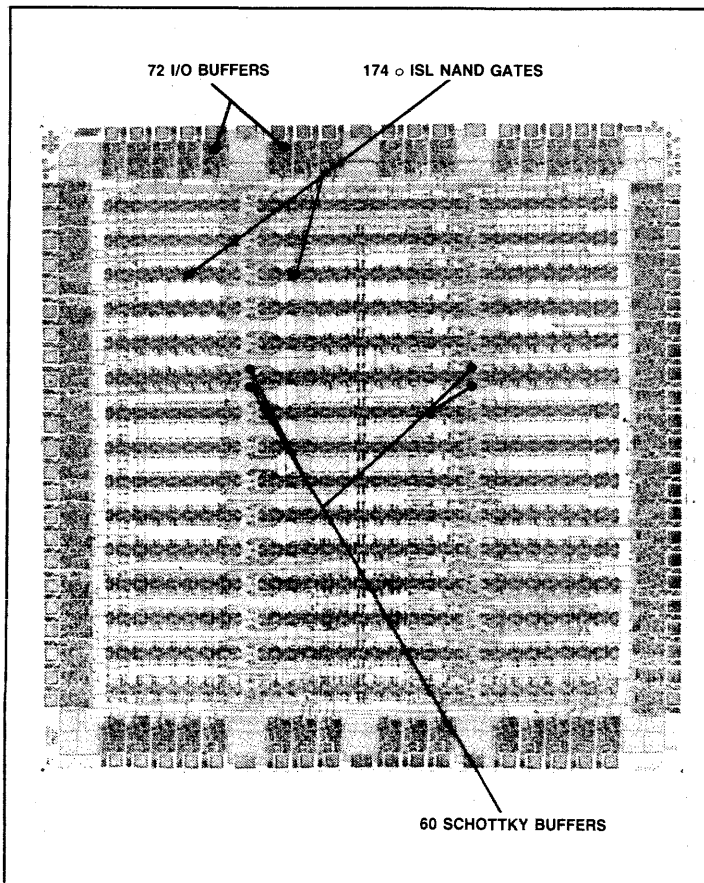


Figure 1. Internal Configuration of 8A1864

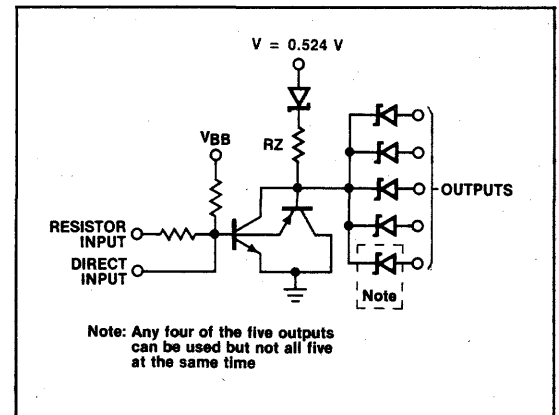


Figure 2. ISL Gate — Schematic Diagram

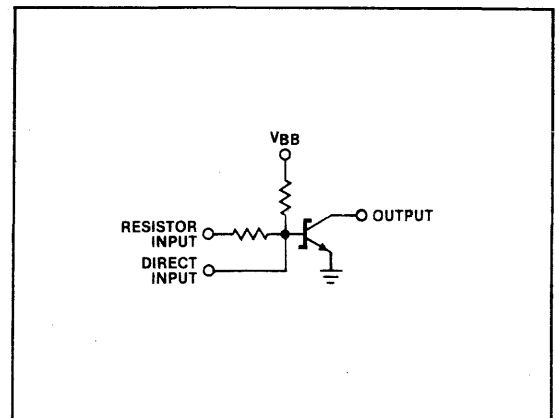


Figure 3. Schottky Buffer — Schematic Diagram

**ISL GATE ARRAY**

**8A2176**

**PRODUCT DESCRIPTION**

The 8A2176 Gate Array (Figure 1) is an uncommitted array of ISL gates (Figure 2), Schottky buffers (Figure 3) and LSTTL-compatible I/O cells. Thus, up to 2016 gates can be custom interconnected to provide the advantages of both Large Scale Integration (LSI) and proprietary design. The 8A2176 76 array is based on a technological subset of LSI called ISL (Integrated Schottky Logic). ISL combines the features of Schottky and the density of I<sup>2</sup>L Bipolar technologies.

Designing with the 8A2176 is easy and fast, requiring no more than conventional logic design, logic simulation, and custom coding of metal interconnections among preprocessed logic gates on the array. The design techniques and the implementation processes are analogous to the design of a Printed Circuit Board.

Logic functions are defined by the user and are implemented by interconnecting 2016 ISL NAND gates, using two layers of metal routing. Seventy-two Schottky buffers are provided to drive multi-load internal clock or enable signals. For external interface, up to 76 LSTTL I/O buffers can be specified.

Each 8-milliampere I/O site can be configured as 1-of-6 input/internal buffers or as 1-of-8 output buffers; each 24-milliampere I/O site can also be configured as 1-of-6 input/internal buffers but the output buffer configuration can be 1-of-12. For a transceiver, either I/O site can be connected in combinations of one input and one output buffer.

**FEATURES**

- Customer programmable LSI
- 2016 ISL (NAND) gates
- 72 Schottky buffers
- 76 I/O buffers
- LS TTL compatible
- Standard PNP inputs
- 8 mA and 24 mA output current sink
- -55°C to +125°C ambient temperature
- 4 ns gate speed (typical)
- Speed power product -0.7 picojoules
- 28, 40, 68, or 84 pin package

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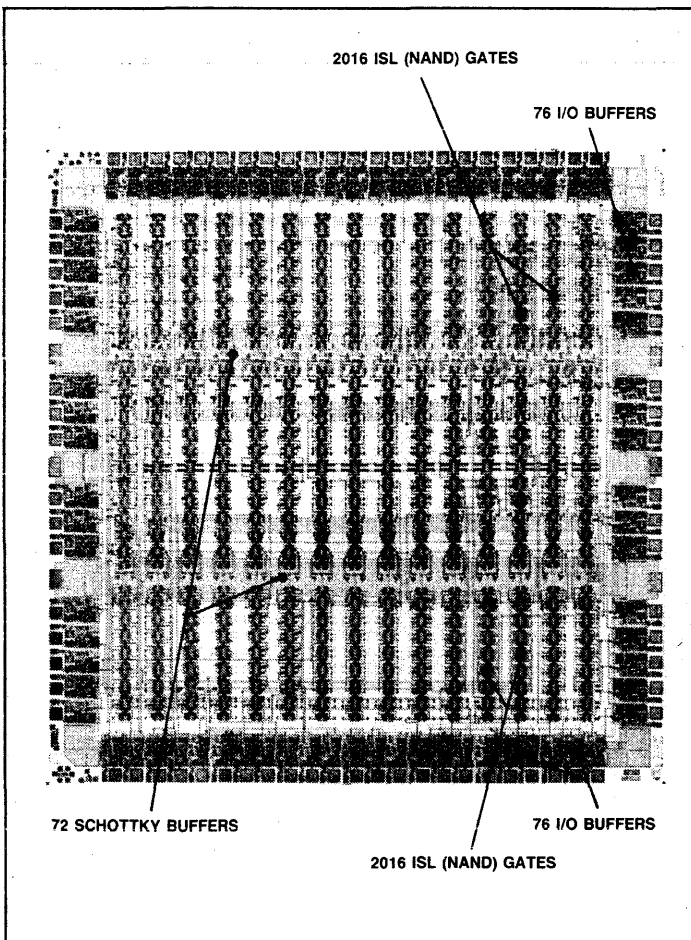


Figure 1. Internal Configuration of 8A2176

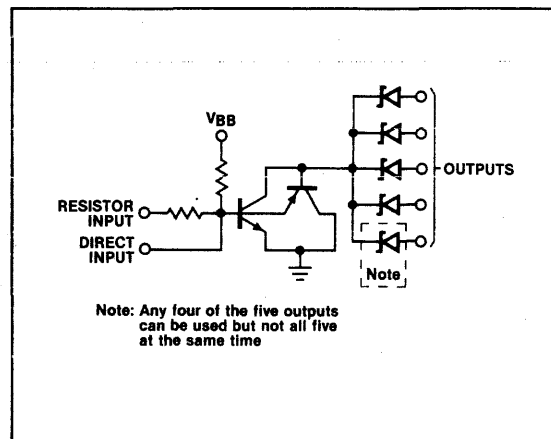


Figure 2. ISL Gate — Schematic Diagram

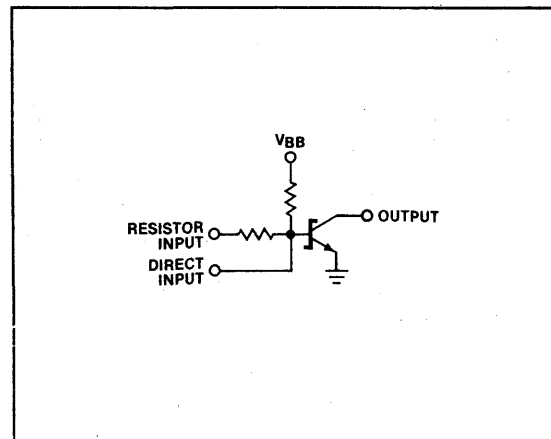


Figure 3. Schottky Buffer — Schematic Diagram

COMPOSITE CELL LOGIC (CCL)

SEMI-CUSTOM FAMILY

PRODUCT DESCRIPTION

Composite Cell Logic (CCL) provides a standard-cell approach to semi-custom bipolar logic. Besides the inherent advantages of LSI and proprietary design, CCL offers the designer a fast turnaround time, a high probability of first-pass success, and a die size that exactly meets all functional requirements of the logic. The CCL approach is particularly well suited to design applications where circuit complexities fall within a range of 100-to-1000 gates.

Figure 1 shows the CCL device together with two standard cells that might be used in the design process. At present, the available cells form two libraries—the Extended Performance Library (EPL) and the Integrated Schottky Library (ISL). Typically, the EPL cell (Figure 2) is used where speed is a critical factor—the speed of EPL cells is comparable to that of Schottky T<sup>2</sup>L logic. *Note. Refer to table elsewhere in this data sheet for nominal figures pertaining to circuit propagation speeds of Schottky, Low-Power Schottky, T<sup>2</sup>L, and Low-Power T<sup>2</sup>L.* All EPL cells are input-expandable with no added delay, are highly immune to internal/external noise, and use active pullups to reduce sensitivity to lead capacitance and the effects of wire-ANDing.

The packing density of an ISL cell (Figure 3) is two to three times greater than that for EPL and the power required is only one-tenth (1/10) to one-twentieth (1/20) as great. The speed of ISL is slightly faster than that of Low-Power Schottky logic. For some circuits, the propagation speeds for ISL and EPL are nearly the same; for other circuits, there are appreciable differences. The speed-comparison table shown later in this data sheet provides a worthwhile guide for overall circuit design.

Output cells of both libraries can sink up to 80-milliamperes of current and both EPL and ISL cells use a 16-micron grid for easy conversion to "Automatic Place and Route" techniques—see Table 1 for a technical summary of both libraries.

Designing with CCL requires a cooperative effort between Signetics and the Customer. The contribution of each party and the overall development sequence are shown in Figure 4.

DESIGN FEATURES

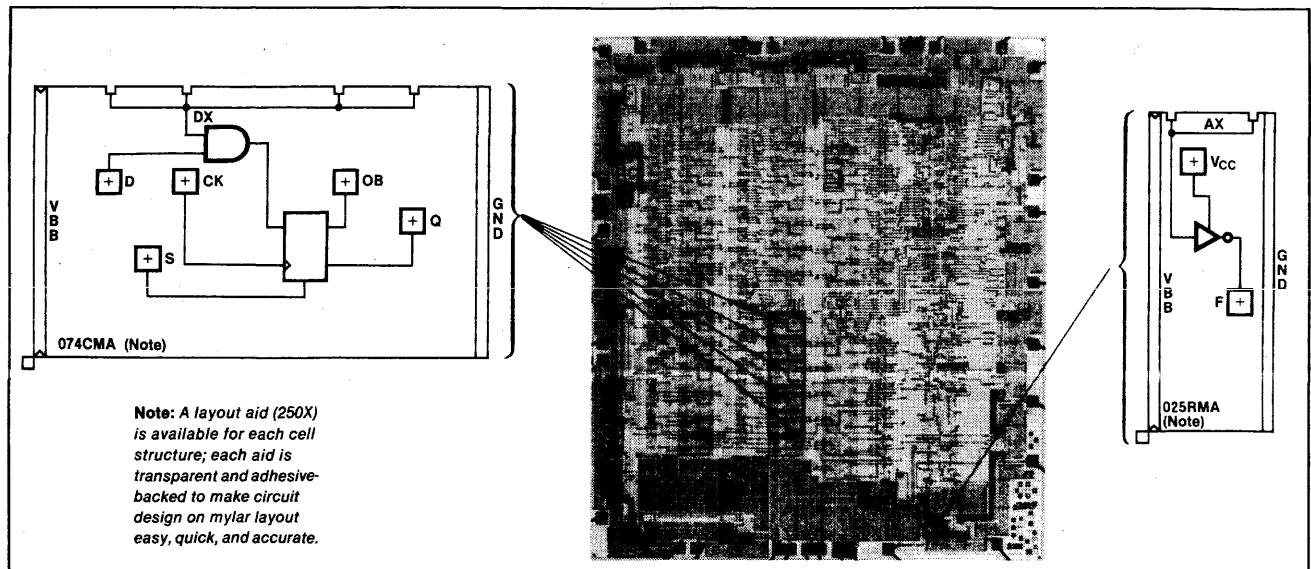
- Customer designed LSI
- Two cell libraries—EPL and ISL
- TTL compatible—each cell is functionally similar to the equivalent 7400 Logic Device
- Two-layer metal interconnects
- PNP or diode inputs
- Open-collector, active pullup, or three-state outputs
- 80-milliampere sink-current capability for output cells
- Accommodate custom cell design
- Most standard packages available
- -55°C to +150°C junction temperature
- +5V (±10%) supply voltage; conditions permitting, on-chip derivation of V<sub>BB</sub> (+1.5V)

Table 1. Technical Summary of EPL and ISL Libraries

PARAMETER	EPL		ISL
	MEDIUM POWER	LOW POWER	
Output structure	Active pullup		Open collector
Input structure	Schottky diode		Schottky diode
Worst-case noise margin	300mV (F.O. = 15)		70mV (F.O. = 15)
Junction temperature range	-55° to +155°C		-55° to +155°C
Power supply	+5V (±10%)		+1.5V (±10%)
Max average speed (in ns)			
F.O. = 1 (T <sub>J</sub> = 150°C)	4.5	5.5	6
F.O. = 6 (T <sub>J</sub> = 150°C)	5.5	7.5	6 or 9*
Max average power (in mW):			
T <sub>J</sub> = 150°C	5.6	2.6	0.3
Packing density gates/mm**	14 to 42		26 to 78

\*Average speed of 6ns requires the use of a resistor pullup cell (optional).

\*\*See Note 5 in Selection Guide regarding derivation of maximum values.



Note: A layout aid (250X) is available for each cell structure; each aid is transparent and adhesive-backed to make circuit design on mylar layout easy, quick, and accurate.

Figure 1. Composite Cell Logic Showing Typical Cell Placement with the Use of Layout Aids.



**COMPOSITE CELL LOGIC (CCL)**

**SEMI-CUSTOM FAMILY**

**ABSOLUTE MAXIMUM RATINGS**

PARAMETER	DESCRIPTION	RATING	UNIT
V <sub>CC</sub>	Supply voltage	+7.0	V
V <sub>BB</sub>	ISL gate supply voltage	+7.0	V
E <sub>IN</sub>	Input voltage, continuous	-0.5 to +5.5	V
I <sub>IN</sub>	Input current, continuous	-30 to +1.0	mA
V <sub>O</sub>	Voltage applied to open-collector output in off-state	-0.5 to +7.0	V
T <sub>A</sub>	Ambient temperature, operating	-55 to +125	°C
T <sub>STG</sub>	Storage temperature	-65 to +150	°C

**AC AND DC ELECTRICAL CHARACTERISTICS**

Conditions<sup>1</sup>: V<sub>CC</sub> = 5.0V (± 10%)  
 V<sub>BB</sub> = 1.5V (± 10%)  
 T<sub>J</sub> = 0°C to +150°C

PARAMETER	TEST CONDITIONS <sup>2</sup>	DESIGN LIMITS <sup>3</sup>			UNIT	
		MIN	TYP	MAX		
<b>EPL GATE (INTERNAL)</b>						
I <sub>CC/G</sub>	Power supply current per gate	0.18	0.29	0.47	mA	
I <sub>LF</sub>	Input load factor		1		UL	
F <sub>O</sub>	Fan-Out		15			
t <sub>pdAV</sub>	Average gate propagation delay = $\frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = 1 EPL gate; Fan-out = 1 EPL gate		3	5.5	ns
t <sub>pdLH</sub>	Propagation delay from low-to-high state	Fan-in = 1 EPL gate; Fan-out = 1 EPL gate		4	7	ns
t <sub>pdHL</sub>	Propagation delay from high-to-low state			2	4	ns
<b>ISL GATE (INTERNAL)</b>						
I <sub>BB/G</sub>	Power supply current per gate	110	150	190	μA	
I <sub>LF</sub>	Input load factor		1		UL	
F <sub>O</sub>	Fan-out		6		UL	
t <sub>pdAV</sub>	Average gate propagation delay = $\frac{t_{pdLH} + t_{pdHL}}{2}$	Fan-in = 1 ISL gate; Fan-out = 1 ISL gate		3	6	ns
t <sub>pdLH</sub>	Propagation delay from low-to-high state	Fan-in = 1 ISL gate; Fan-out = 1 ISL gate		5	10	ns
t <sub>pdHL</sub>	Propagation delay from high-to-low state			1	2	ns
<b>EPL/ISL INPUT CELLS</b>						
V <sub>TH</sub>	Input threshold voltage	0.8		2.0	V	
V <sub>CD</sub>	Input clamp diode voltage	I <sub>IN</sub> = -18mA		-1.2	V	
I <sub>IL</sub>	Input low current	V <sub>IN</sub> = 0.4V	PNP input	-20	mA	
			Diode input	-400		
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = 2.7V		20	μA	
I <sub>I</sub>	Maximum input high current	V <sub>IN</sub> = 5.5V		100		
F <sub>O</sub>	Fan-out (ISL library)	Standard cell		6	UL	
		Clock buffer cell		64		
	Fan-out (EPL library)	Standard cell		15		
		Clock buffer cell		55		
	Clock buffer cell		100			

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## COMPOSITE CELL LOGIC (CCL)

## SEMI-CUSTOM FAMILY

## AC AND DC ELECTRICAL CHARACTERISTICS (Continued)

PARAMETER	TEST CONDITIONS <sup>2</sup>	DESIGN LIMITS <sup>3</sup>			UNIT	
		MIN	TYP	MAX		
<b>EPL/ISL THREE-STATE OUTPUT BUFFERS</b>						
$I_{LF}$ Input load factor	EPL			1 or 9	UL	
	ISL			2 or 4		
$V_{OL}$ Output low voltage	Military: 4mA			400	mV	
	Commercial: 8mA			500		
	Military: 12 mA			400	mV	
	Commercial: 24 mA			500		
$V_{OH}$ Output high voltage	Military: $I_{OH} = -400 \mu A$	2.5			V	
	Commercial: $I_{OH} = -400 \mu A$	2.7				
$I_{OS}$ Output short circuit current	$V_{OUT} = 0V$	-15		-100	mA	
$I_{OLZ}$ Three-state off current (output low)	$V_{OUT} = 0.4V$			-20	$\mu A$	
$I_{OHZ}$ Three-state off current (output high)	$V_{OUT} = 2.4V$			20	$\mu A$	
<b>EPL/ISL OPEN-COLLECTOR OUTPUT BUFFERS</b>						
$I_{LF}$ Input load factor	EPL	8 mA output			1	UL
		20 mA output			3	
		80 mA output			5	
	ISL			2 or 4		
$V_{OL}$ Output low voltage	$I_{OL}$	8 mA output			500	mV
		20 mA			500	
		70 mA			500	
		80 mA			800	
$I_{OH}$ Output leakage current	$V_{OUT}$	80 mA Cells	2.75		60	$\mu A$
			5.5V		250	
		8/20 mA Cells	5.5V		100	
<b>ACTIVE PULLUP OUTPUT BUFFER</b>						
$I_{LF}$ Input load factor	EPL			1 or 3	UL	
	ISL			2 or 4		
$V_{OL}$ Output low voltage	$I_{OL} = 8 \text{ mA or } 20 \text{ mA}$			500	mV	
$V_{OH}$ Output high voltage	Military $I_{OH} = -400 \mu A$ or $-1.0 \text{ mA}$	2.5			V	
	Commercial $I_{OH} = -400 \mu A$ or $-1.0 \text{ mA}$	2.7				
$I_{OS}$ Output short circuit current	$V_{OUT} = 0V$	8 mA output	-15		-100	mA
		20 mA output	-40		-100	

## Notes:

- Maximum power dissipation is determined from individual cell data sheets; the figures are then summed to calculate total power for the chip. The total power must be less than the Maximum Power Dissipation ( $P_d$ ) calculated earlier in this data sheet.
- For test circuits and timing waveforms, refer to individual cell data sheets in the CCL Design Guide.
- Design limits are based on standard modeling with similar circuits used in Signetics standard bipolar LSI products. Actual simulation limits are maintained to be consistent with characterization updates for the CCL libraries.

## ORDERING INFORMATION

For additional information pertaining to Composite Cell Logic, Gate Arrays, and other Semicustom Products, contact the nearest Signetics Sales Office.

**ADVANCED CUSTOMIZED ECL (ACE)**

**ACE 600 & ACE 900**

**MASTER SLICE LOGIC ARRAYS**

**PRODUCT DESCRIPTION**

The Signetics Advanced Customized ECL (ACE) family of products provides the user with a cost-effective technology, futuristic speeds, and other high-performance alternatives for the design of LSI-based systems. Basic cell designs are implemented with Emitter Coupled and Common Mode Logic (ECL/CML) to guarantee the very best compromise between speed, power, and interface capabilities—see Figure 1 and TECHNICAL SUMMARY that follows.

At present, the ACE product line is available with gate complements of 600, 900, 1320, 1400, and 2200; the 1320 array actually contains 1000 gates with an on-board 320-bit RAM. The 600/900-gate arrays, described in this data sheet, are well-suited for low-cost applications and for use in systems that do not require 25-ohm terminations. To meet the flexibility requirements, the rise-and-fall times for I/O cells of these arrays are mask-selectable and bidirectional and TTL interfaces are standard.

All ACE arrays are I/O compatible with the 10K/100K ECL logic family and all are fabricated with a very mature process; thus, even with 3-micron first-metal geometry, first pass success is a virtual certainty. The speed-power product for devices in the ACE family is in the neighborhood of 1 to 3 picojoules, permitting heat-sink cooling at ambient air

temperatures. The ACE family and MACRO library is alternately sourced by a major supplier of semicustom devices.

To summarize, the designer, using ACE, is limited only by innovation and imagination:

- ECL/CML Technology for SPEED and EFFICIENCY
- Mature process for PRODUCT CERTAINTY
- Computer aided design for QUICK DELIVERY
- Pin grid packages (socket insertion) for RELIABILITY
- Signetics for QUALITY

**FEATURES**

- 3-micron geometry (first metal)
- Internal gate delays as low as 300 picoseconds (average gate delay of 450 picoseconds)
- Expandable 80-cell MACRO library
- Mask-selectable rise/fall times for I/O interface cells
- Bidirectional and TTL interfaces
- 10K/100K ECL compatibility
- Computer aided design (CAD) for layout, simulation, and testing
- Mature process (SUBILO P)
- Pin grid array packages for easy socket insertion
- 25 and 50 ohm drive capability

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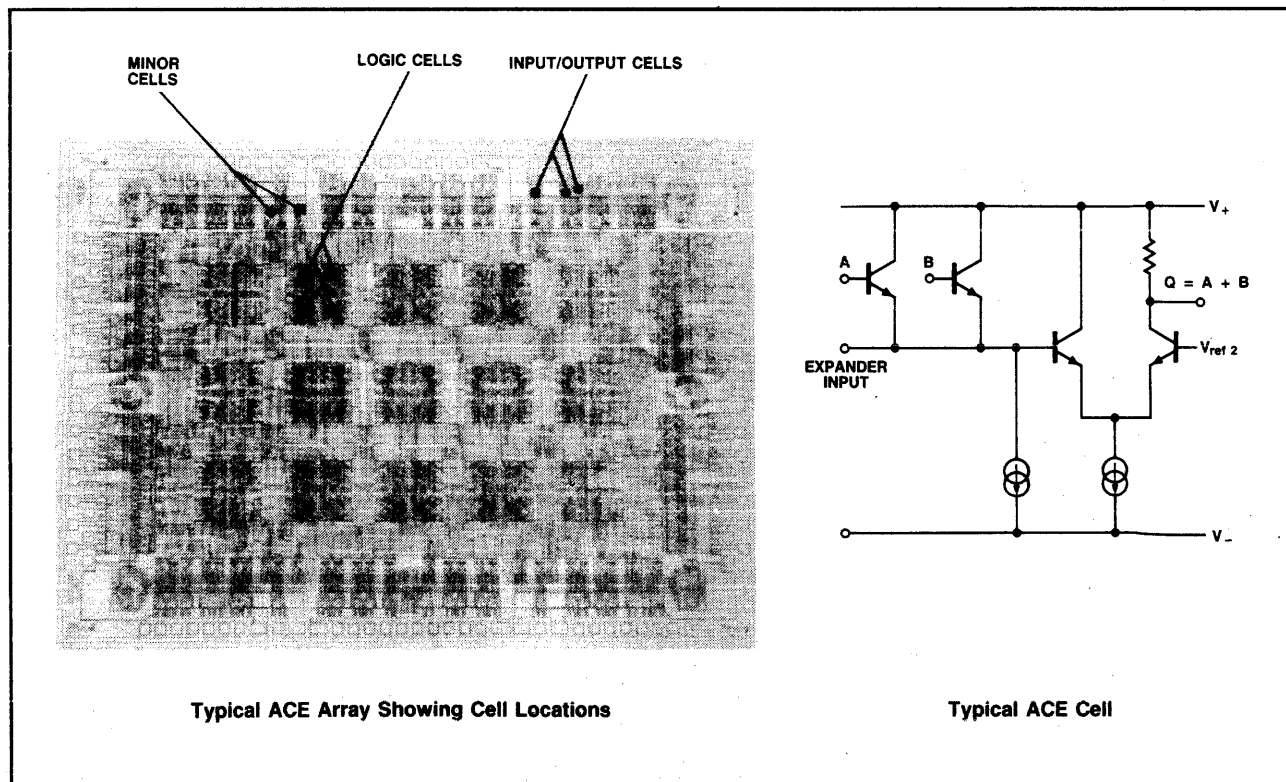


Figure 1. Chip Architecture and Typical Circuit

**ADVANCED CUSTOMIZED ECL (ACE)**
**ACE 600 & ACE 900**
**MASTER SLICE LOGIC ARRAYS**
**TECHNICAL SUMMARY OF ACE FAMILY**

PARAMETER	ACE 600	ACE 900	ACE 1320	ACE 1400	ACE 2200
Major cells	24	36	48	60	100
Input/output cells	28	28	96	96	128
Input cells	30	42	—	—	—
Worst case noise margin	24 - 45 mV		24 - 45 mV		
Junction temperature range	30 - 125°C		30 - 125°C		
Average prop delay (internal gate)	0.3 - 0.5 ns		0.3 - 0.5 ns		
	<b>10K LEVEL</b>	<b>100K LEVEL</b>	<b>10K LEVEL</b>	<b>100K LEVEL</b>	
Power supply	-5.25V ± 5%	-4.5V ± 5%	-5.25V ± 5%	-4.5V ± 5%	
Power consumption	2.1 - 2.7 mW	1.8 - 2.3 mW	4.6 - 6.3 mW	4 - 5.5 mW	
<b>ACE PACKAGE TYPE AND THERMAL RESISTIVITY SELECTION</b>					
	<b>PACKAGE</b>	<b>THERMAL RESISTIVITY (°C/W)</b>			
		<b>HEAT SINK</b>	<b>NO AIR FLOW</b>	<b>5 m/s AIR FLOW</b>	
ACE 600 & ACE 900	64 Pin	Yes	25	13	
		No	50	25	
ACE 1320, ACE 1400 & ACE 2200	144 Pin	Yes	12	6	
		No	24	12	

# ADVANCED CUSTOMIZED ECL (ACE)

# ACE 1320, ACE 1400 & ACE 2200

## MASTER SLICE LOGIC ARRAYS

### PRODUCT DESCRIPTION

The Signetics Advanced Customized ECL (ACE) family of products provides the user with a cost-effective technology, futuristic speeds, and other high-performance alternatives for the design of LSI-based systems. Basic cell designs are implemented with Emitter Coupled and Common Mode Logic (ECL/CML) to guarantee the very best compromise between speed, power, and interface capabilities—see Figure 1 and TECHNICAL SUMMARY that follows.

At present, the ACE product line is available with gate complements of 600, 900, 1320, 1400, and 2200; the 1320 array actually contains 1000 gates with an on-board 320-bit RAM. The 1320/1400/2200 gate arrays, described in this data sheet, are particularly well suited for complex applications requiring relatively high gate counts and considerable design flexibility. To meet the flexibility requirements, the rise-and-fall times for I/C cells of these arrays are mask-selectable and bidirectional and TTL interfaces are standard.

All ACE arrays are I/O compatible with the 10K/100K ECL logic family and all are fabricated with a very mature process; thus, even with 3-micron first-metal geometry, first pass success is a virtual certainty. The speed-power product for devices in the ACE family is in the neighborhood of 1 to 3 picojoules, permitting heat-sink cooling at ambient air

temperatures. The ACE family and MACRO library is alternately sourced by a major supplier of semicustom devices.

To summarize, the designer, using ACE, is limited only by innovation and imagination:

- ECL/CML Technology for SPEED and EFFICIENCY
- Mature process for PRODUCT CERTAINTY
- Computer aided design for QUICK DELIVERY
- Pin grid packages (socket insertion) for RELIABILITY
- Signetics for QUALITY

### FEATURES

- 3-micron geometry (first metal)
- Internal gate delays as low as 300 picoseconds (average gate delay of 450 picoseconds)
- Expandable 80-cell MACRO library
- Mask-selectable rise/fall times for I/O interface cells
- Bidirectional and TTL interfaces
- 10K/100K ECL compatibility
- Computer aided design (CAD) for layout, simulation, and testing
- Mature process (SUBILO P)
- Pin grid array packages for easy socket insertion
- 25-ohm and 50-ohm load drive capability

Signetics

CUSTOM/SEMICUSTOM

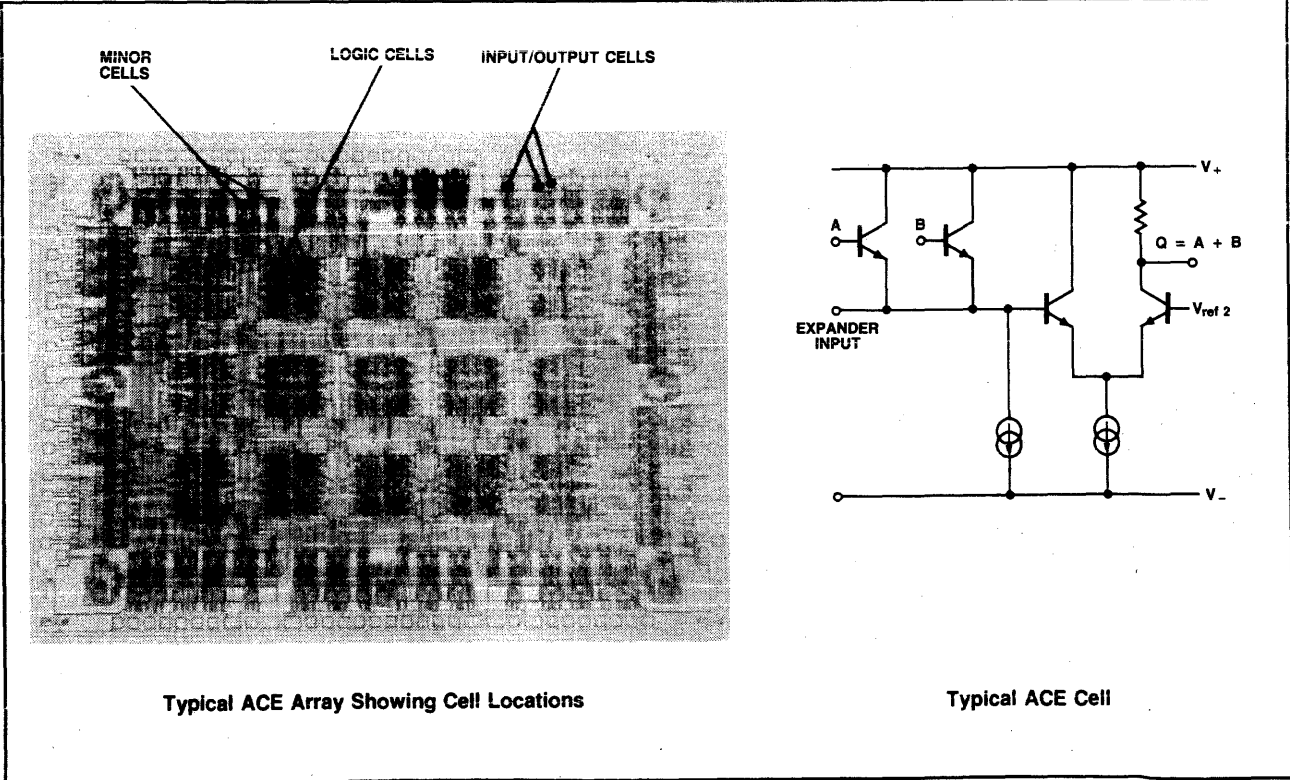


Figure 1. Chip Architecture and Typical Circuit

**ADVANCED CUSTOMIZED ECL (ACE)**
**ACE 1320, ACE 1400 & ACE 2200**

MASTER SLICE LOGIC ARRAYS

**TECHNICAL SUMMARY OF ACE FAMILY**

PARAMETER	ACE 600	ACE 900	ACE 1320	ACE 1400	ACE 2200
Major cells	24	36	48	60	100
Input/output cells	28	28	96	96	128
Input cells	30	42	—	—	—
Worst case noise margin	24 - 45 mV		24 - 45 mV		
Junction temperature range	30 - 125°C		30 - 125°C		
Average prop delay (internal gate)	0.3 - 0.5 ns		0.3 - 0.5 ns		
	<b>10K LEVEL</b>	<b>100K LEVEL</b>	<b>10K LEVEL</b>	<b>100K LEVEL</b>	
Power supply	-5.25V ± 5%	-4.5V ± 5%	-5.25V ± 5%	-4.5V ± 5%	
Power consumption	2.1 - 2.7 mW	1.8 - 2.3 mW	4.6 - 6.3 mW	4 - 5.5 mW	
<b>ACE PACKAGE TYPE AND THERMAL RESISTIVITY SELECTION</b>					
	<b>PACKAGE</b>	<b>THERMAL RESISTIVITY (°C/W)</b>			
		<b>HEAT SINK</b>	<b>NO AIR FLOW</b>	<b>5 m/s AIR FLOW</b>	
ACE 600 & ACE 900	64 Pin	Yes	25	13	
		No	50	25	
ACE 1320, ACE 1400 & ACE 2200	144 Pin	Yes	12	6	
		No	24	12	

# CMOS GATE ARRAYS (M-Series)

SCC0330-M, SCC0450-M,  
SCC0700-M, SCC1100-M

## PRODUCT DESCRIPTION

The SCCXXXX gate array family offers the circuit designer the facility to create a semi-custom circuit with a unique set of CAD (Computer-Aided Design) tools in a well-established CMOS process.

Signetics M-Series CMOS Gate Arrays are single chip programmable devices that allow customization of user logic. Only metalization and contact are programmed in these mature CMOS devices. Thus, fast turnaround from logic to completed silicon is achieved.

Each device in this family of low power gate arrays contains numerous identical, uncommitted unit cells (Figure 1) which are interconnected by two custom masks (metal and contact). Each unit cell contains four pairs of N and P transistors. Access to the transistors is from both the top and bottom of the cells and, additionally, there are two poly feed-throughs at each side of the cell. This homogenous cell design allows for excellent routing flexibility, and many designs result in better than 80% utilization of the gates available.

The M-Series Gate Arrays are built on a mature, state-of-the-art 4-micron Si-gate CMOS process incorporating an epi-substrate, which significantly reduces the potential for latch as compared with other bulk CMOS processes.

Computer Aided Design (CAD) is used throughout the design process to ensure accurate implementation of customer logic (see Figure 14 for typical process flow).

## FEATURES

- Customer programmable LSI
- 330 to 1100 gate complexity
- Mature silicon gate technology with local oxidation
- Library of 60 pre-designed, fully characterized macrocells available
- Full CAD, including auto-place and auto-route, for quick error-free design
- Very low power consumption (e.g. standby power for SCC 0700 is 0.25mW)
- Excellent noise immunity
- Power supply range 3 to 15V
- Over 80% utilization typical
- Fully programmable I/O pins, each having a wide range of functions
- Input protection by series resistor and diode clamp to V<sub>SS</sub>
- TTL outputs (buffers) drive up to four LSTTL loads
- -55°C to +125°C operating temperature
- Plastic and ceramic DIP, ceramic leadless chip carriers, and plastic leaded chip carriers available

Signetics

CUSTOM/SEMICUSTOM

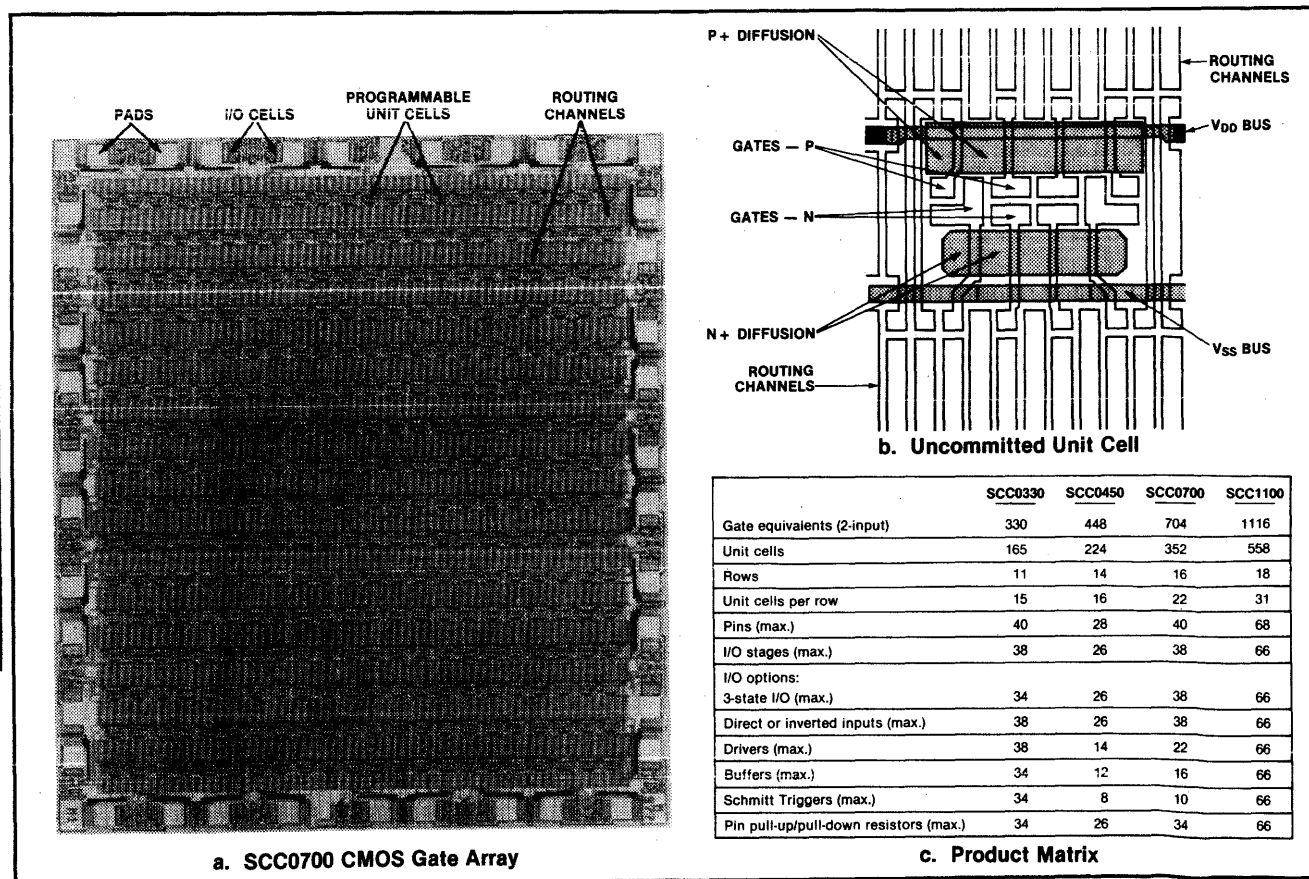


Figure 1. Internal Configuration and Functional Characteristics of M-Series CMOS Gate Arrays

# CMOS GATE ARRAYS (M-Series)

SCC0330-M, SCC0450-M,  
SCC0700-M, SCC1100-M

## DC ELECTRICAL CHARACTERISTICS (Continued) $V_{SS}=0V$ ; for all devices unless otherwise specified.

Parameter	Operating Temp ( $T_{amb}$ ) <sup>1</sup>	Supply Voltage	Conditions	Temperature Range <sup>1</sup>						Unit
				$T_{amb} = \text{Low}$		$T_{amb} = 25^\circ\text{C}$		$T_{amb} = \text{High}$		
				Min	Max	Min	Max	Min	Max	
$I_{OL}$ Output (sink) current Low driver outputs	Standard	5	$V_I = 0V \text{ or } 5V;$ $V_O = 0.4V$	1.1	—	0.9	—	0.7	—	mA
		10		4.0	—	3.3	—	2.6	—	
		15		12.0	—	10.0	—	8.0	—	
	Extended	5		1.2	—	0.9	—	0.6	—	
		10		4.2	—	3.3	—	2.2	—	
		15		13.0	—	10.0	—	6.7	—	
$I_{OL}$ Output (sink) current Low buffer outputs	Standard	5	$V_I = 0V \text{ or } 10V;$ $V_O = 0.5V$	2.2	—	1.8	—	1.4	—	
		10		8.0	—	6.6	—	5.6	—	
		15		24.0	—	20.0	—	16.0	—	
	Extended	5		2.4	—	1.8	—	1.2	—	
		10		8.4	—	6.6	—	4.4	—	
		15		26.0	—	20.0	—	13.4	—	
$-I_{OH}$ Output (source) current High	Standard	5	$V_I = 0V \text{ or } 5V;$ $V_O = 4.6V$	1.1	—	0.9	—	0.7	—	
		10		3.1	—	2.6	—	2.0	—	
		15		12.0	—	10.0	—	8.0	—	
	Extended	5		1.2	—	0.9	—	0.6	—	
		10		3.5	—	2.6	—	1.7	—	
		15		13.0	—	10.0	—	6.7	—	
$\pm I_{IN}$ Input leakage current	Standard	10	$V_I = 0V \text{ or } 10V$	—	0.3	—	0.3	—	1.0	
		15		—	0.3	—	0.3	—	1.0	
	Extended	10	$V_I = 0V \text{ or } 15V$	—	0.1	—	0.1	—	1.0	
		15		—	0.1	—	0.1	—	1.0	
$I_{OZH}$ Three-state output and open N-channel output leakage current High	Standard	10	Output returned to $V_{DD}$	—	1.6	—	1.6	—	12.0	
		15		—	1.6	—	1.6	—	12.0	
	Extended	10		—	0.4	—	0.4	—	5.0	
		15		—	0.4	—	0.4	—	5.0	
$-I_{OZL}$ Three-state output and open P-channel output leakage current Low	Standard	10	Output returned to $V_{SS}$	—	1.6	—	1.6	—	12.0	
		15		—	1.6	—	1.6	—	12.0	
	Extended	10		—	0.4	—	0.4	—	5.0	
		15		—	0.4	—	0.4	—	5.0	
$V_{TH}$ Upper threshold voltage	Standard	5	Internal Schmitt trigger	—	—	3.4	Typical values	—	—	
		10		—	—	6.8		—	—	
		15		—	—	10.2		—	—	
$V_{TL}$ Lower threshold voltage		5		—	—	2.2	Typical values	—	—	
		10		—	—	3.0		—	—	
		15		—	—	3.8		—	—	
$V_H$ Hysteresis voltage input: INPS	—	5	—	—	—	0.2	Typical values	—	—	
		10		—	—	0.6		—	—	
		15		—	—	0.8		—	—	

NOTES:  
 1.  $T_{amb}$  Low:  $-40^\circ\text{C}$  for standard temperature range  $T_{amb}$  High:  $+85^\circ\text{C}$  for standard temperature range  
 $-55^\circ\text{C}$  for extended temperature range  $+125^\circ\text{C}$  for extended temperature range  
 2. Pin-connected pull-up and pull-down resistors are typically 7 to 78 K-ohms — see PERIPHERY.  
 3. When pull-up or pull-down resistors are used, current limits for  $I_{DD}$  must be extrapolated.



# CMOS GATE ARRAYS (M-Series)

SCC0330-M, SCC0450-M,  
SCC0700-M, SCC1100-M

## DC CHARACTERISTICS (Continued)

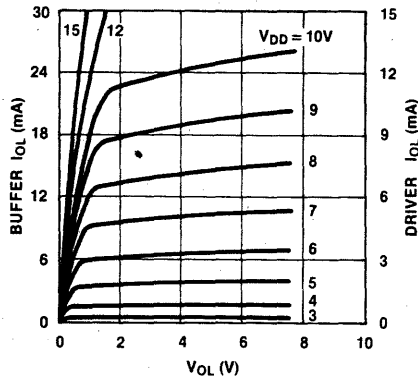


Figure 2. Minimum Output Current LOW as a Function of the Output Voltage LOW; Buffer and Driver Outputs

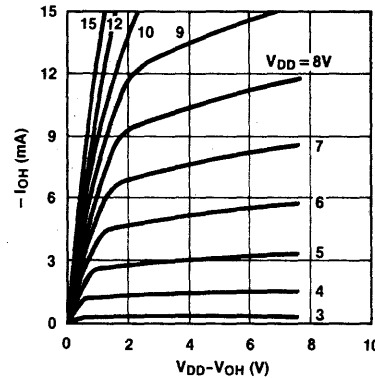


Figure 3. Minimum Output Current HIGH as a Function of the Supply Voltage Minus the Output Voltage HIGH

## AC ELECTRICAL CHARACTERISTICS $V_{SS} = 0V; T_{amb} = 25^{\circ}C$

Parameter	Pwr Sup ( $V_{DD}$ )	Min	Typ	Max	Unit	Parameter	Pwr Sup ( $V_{DD}$ )	Min	Typ	Max	Unit
$f_{max}$ Maximum toggle frequency flip-flop GT00 (no set/reset)	5	6	12	—	MHz	OUTPUT STAGE TRANSITION TIMES: Input transition $\leq 20ns, C_L = 50pF; V_{SS} = 0V, T_{amb} = 25^{\circ}C$					
	10	12	24	—	MHz						
	15	15	30	—	MHz						
$f_s$ Maximum system frequency (may depend on number of gates in sequence)	5	3	6	—	MHz	$t_{THL}$ Driver outputs High-to-Low	5	60	120	ns	
	10	6	12	—	MHz		10	30	60	ns	
	15	9	18	—	MHz	15	20	40	ns		
$t_p$ Propagation delays for 2-input NAND gate with fanout of 2	5	—	8	16	ns	$t_{THL}$ Buffer outputs High-to-Low	5	30	60	ns	
	10	—	3.2	6.4	ns		10	15	30	ns	
	15	—	2	4	ns	15	10	20	ns		
						$t_{TLH}$ Buffer outputs Low-to-High	5	40	80	ns	
						10	18	36	ns		
						15	12	24	ns		

## GATE DELAYS

### Nominal Propagation Delay

In Figures 6 through 12, examples are given of the nominal propagation delay times of several library cells, these being calculated from the delay figures given in the individual macro descriptions. These graphs are intended to provide quick-reference data to enable the designer to make an esti-

mate of critical a.c. path without having built or simulated a network.

Accurate delay figures can only be obtained after incorporating the wiring length load automatically calculated by INGATE (i.e., the result of the automatic routing program). A maximum delay is obtained by multiplying the nominal value by 2.2.

Signetics CUSTOM/SEMICUSTOM

# CMOS GATE ARRAYS (M-Series)

SCC0330-M, SCC0450-M,  
SCC0700-M, SCC1100-M

## ABSOLUTE MAXIMUM RATINGS Limiting values in accordance with the Absolute Maximum System (IEC 134)

Supply voltage range	$V_{DD}$	-0.5 to +18V
Voltage on any input when pin pull up/down resistors are: Used Not used	$V_I$ $V_I$	-0.5 to $V_{DD} + 0.5V$ -0.5 to +18V
DC current into any input or output	$\pm 1$	Max. 10mA
Power dissipation per output	P	Max. 100mW
Power dissipation per package For standard temperature range: -40°C to +85°C (plastic and ceramic DIP) For $T_{amb} = -40^\circ C$ to +60°C For $T_{amb} = +60^\circ C$ to +85°C	$P_{tot}$ $P_{tot}$	Max. 400mW Derate linearly by 8mW/K to 200mW
For extended temperature range: -55°C to +125°C (ceramic DIP) For $T_{amb} = -55^\circ C$ to +100°C For $T_{amb} = +100^\circ C$ to +125°C	$P_{tot}$ $P_{tot}$	Max. 400mW Derate linearly by 8mW/K to 200mW
Storage temperature range	$T_{stg}$	-65°C to +150°C

## DC ELECTRICAL CHARACTERISTICS $V_{SS} = 0V$ ; for all devices unless otherwise specified.

Parameter	Operating Temp ( $T_{amb}$ ) <sup>1</sup>	Supply Voltage	Conditions	Temperature Range <sup>1</sup>						Unit
				$T_{amb} = \text{Low}$		$T_{amb} = 25^\circ C$		$T_{amb} = \text{High}$		
				Min	Max	Min	Max	Min	Max	
$I_{DD}$ Quiescent device current	Standard	5	All valid input combinations; $V_i = V_{SS}$ or $V_{DD}$	—	50	—	50	—	375	$\mu A$
		10		—	100	—	100	—	750	
		15		—	200	—	200	—	1500	
	Extended	5		—	15	—	15	—	375	
		10		—	25	—	25	—	750	
		15		—	50	—	50	—	1500	
$V_{OL}$ Output voltage Low	Both standard and extended ranges	5	$V_i = V_{SS}$ or $V_{DD}$ ; $I_o < 1.0\mu A$	—	0.05	—	0.05	—	0.05	V
		10		—	0.05	—	0.05	—	0.05	
		15		—	0.05	—	0.05	—	0.05	
$V_{OH}$ Output voltage High		5		4.95	—	4.95	—	4.95	—	
		10		9.95	—	9.95	—	9.95	—	
		15		14.95	—	14.95	—	14.95	—	
$V_{IL}$ Input voltage Low: INPI/INPB	Both standard and extended ranges	5	$V_o = 0.5V$ or $4.5V$ ; $I_o < 1.0\mu A$	—	1.5	—	1.5	—	1.5	V
		10		—	3.0	—	3.0	—	3.0	
		15		—	4.0	—	4.0	—	4.0	
$V_{IH}$ Input voltage High: INPI/INPB		5	$V_o = 1.0V$ or $9.0V$ ; $I_o < 1.0\mu A$	3.5	—	3.5	—	3.5	—	
		10		7.0	—	7.0	—	7.0	—	
		15		11.0	—	11.0	—	11.0	—	
$V_{IL}$ Input voltage Low: INPA, INPD, INPS	Both standard and extended ranges	5	$V_o = 0.5V$ or $4.5V$ ; $I_o < 1.0\mu A$	—	1.0	—	1.0	—	1.0	V
		10		—	2.0	—	2.0	—	2.0	
		15		—	2.5	—	2.5	—	2.5	
$V_{IH}$ Input voltage High: INPA, INPD, INPS		5	$V_o = 1.0V$ or $9.0V$ ; $I_o < 1.0\mu A$	4.0	—	4.0	—	4.0	—	
		10		8.0	—	8.0	—	8.0	—	
		15		12.5	—	12.5	—	12.5	—	

**CMOS GATE ARRAYS (H-Series)**

SCC0335-H, SCC0455-H  
SCC0705-H, SCC1105-H

**Preliminary**

**PRODUCT DESCRIPTION**

The SCCXXX gate array family offers the circuit designer the facility to create a semi-custom circuit with a unique set of CAD (Computer-Aided Design) tools in a well-established CMOS process.

Signetics H-Series CMOS Gate Arrays are single chip programmable devices that allow customization of user logic. Only metalization and contact are programmed in these mature CMOS devices. Thus, fast turnaround from logic to completed silicon is achieved.

Each device in this family of low power gate arrays contains numerous identical, uncommitted unit cells (Figure 1) which are interconnected by two custom masks (metal and contact). Each unit cell contains four pairs of N and P transistors. Access to the transistors is from both the top and bottom of the cells and, additionally, there are two poly feed-throughs at each side of the cell. This homogenous cell design allows for excellent routing flexibility, and many designs result in better than 80% utilization of the gates available.

The H-Series Gate Arrays are built on a mature, state-of-the-art 3-micron Si-gate CMOS process incorporating an epi-substrate, which significantly reduces the potential for latch as compared with other bulk CMOS processes.

Computer Aided Design (CAD) is used throughout the design process to ensure accurate implementation of customer logic (see Figure 14 for typical process flow).

**FEATURES**

- Customer programmable LSI
- 330 to 1100 gate complexity
- Mature silicon gate technology with local oxidation
- Library of 60 pre-designed, fully characterized macrocells available
- Full CAD, including auto-place and auto-route, for quick error-free design
- Very low power consumption (e.g. standby power for SCC 0705 is 0.25mW)
- Excellent noise immunity
- Power supply range 3 to 6V
- Over 80% utilization typical
- Fully programmable I/O pins, each having a wide range of functions
- Input protection by series resistor and diode clamp to V<sub>ss</sub>
- TTL outputs (buffers) drive up to four LSTTL loads
- -55°C to +125°C operating temperature
- Plastic and ceramic DIP, ceramic leadless chip carriers, and plastic leaded chip carriers available

Signetics

CUSTOM/SEMICUSTOM

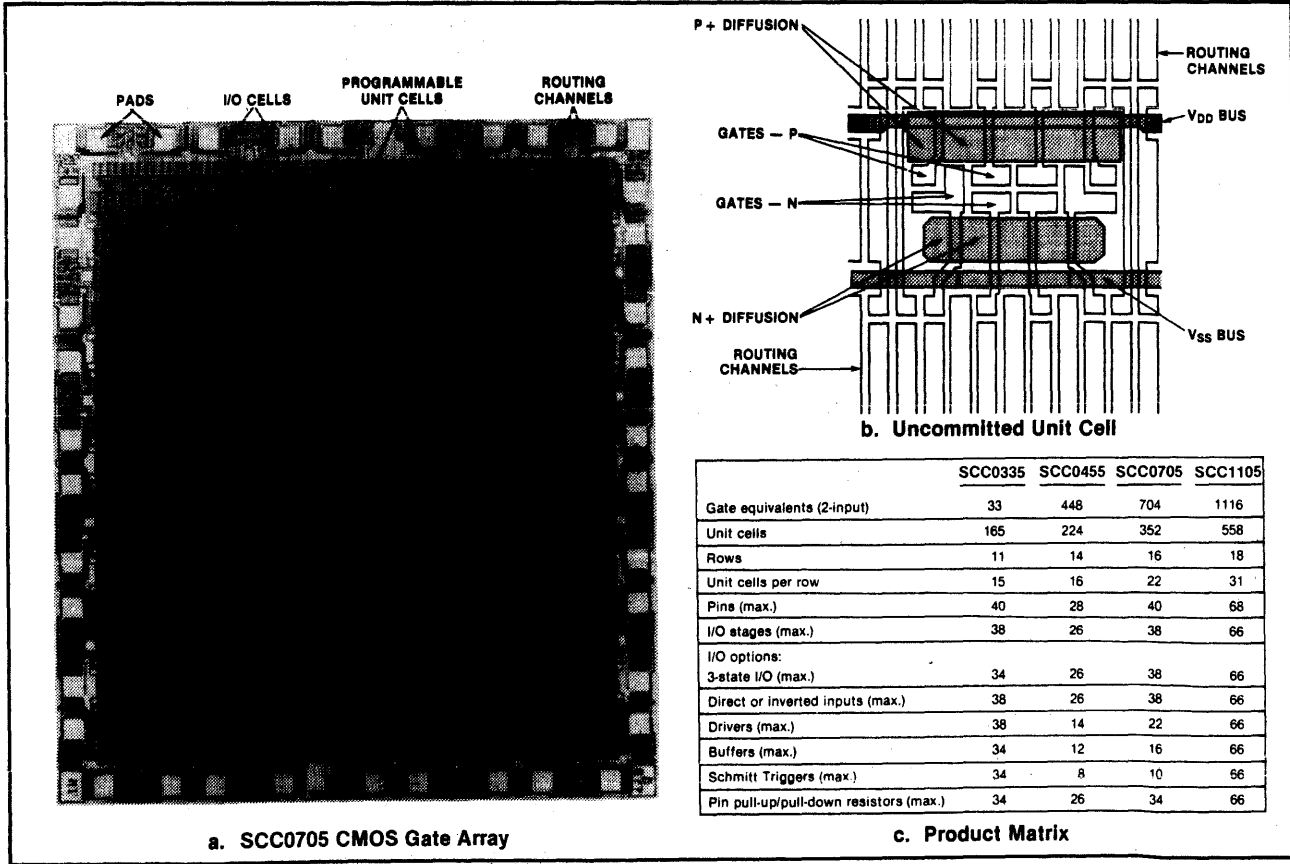


Figure 1. Internal Configuration and Functional Characteristics of H-Series CMOS Gate Arrays

# CMOS GATE ARRAYS (H-Series)

SCC0335-H, SCC0455-H  
SCC0705-H, SCC1105-H

**Preliminary**

## ABSOLUTE MAXIMUM RATINGS

Limiting values in accordance with the Absolute Maximum System (IEC 134)

Supply voltage range	$V_{DD}$	- 0.5 to + 7V
Voltage on any input when pin pull up/down resistors are: Used Not used	$V_I$ $V_I$	- 0.5 to $V_{DD} + 0.5V$ - 0.5 to + 7V
DC current into any input or output	$\pm 1$	Max. 10mA
Power dissipation per output	P	Max. 100mW
Power dissipation per package For standard temperature range: - 40°C to + 85°C (plastic and ceramic DIP) For $T_{amb} = - 40^\circ C$ to + 60°C For $T_{amb} = + 60^\circ C$ to + 85°C	$P_{tot}$ $P_{tot}$	Max. 400mW Derate linearly by 8mW/K to 200mW
For extended temperature range: - 55°C to + 125°C (ceramic DIP) For $T_{amb} = - 55^\circ C$ to + 100°C For $T_{amb} = + 100^\circ C$ to + 125°C	$P_{tot}$ $P_{tot}$	Max. 400mW Derate linearly by 8mW/K to 200mW
Storage temperature range	$T_{stg}$	- 65°C to + 150°C

## DC ELECTRICAL CHARACTERISTICS

$V_{SS} = 0V$ ; for all devices unless otherwise specified.

SYMBOL AND PARAMETER	OPERATING TEMP ( $T_{amb}$ ) <sup>1</sup>	SUPPLY VOLTAGE	TEST CONDITIONS	TEMPERATURE RANGE <sup>1</sup>						UNIT
				$T_{amb} = \text{LOW}$		$T_{amb} = 25^\circ C$		$T_{amb} = \text{HIGH}$		
				MIN	MAX	MIN	MAX	MIN	MAX	
$I_{DD}$ Quiescent device current	Standard	5	$I_O = 0$ $V_{CC}$ or GND	—	50	—	50	—	375	$\mu A$
		—		—	—	—	—	—		
		—		—	—	—	—	—		
	Extended	5		—	15	—	15	—	375	
		—		—	—	—	—	—	—	
		—		—	—	—	—	—	—	
$V_{OL}$ Output voltage Low	Both standard and extended ranges	2.0	$V_{IH}$ or $V_{IL}$ $I_O = 20\mu A$	—	0.1	—	0.1	—	0.1	V
		4.5		—	0.1	—	0.1	—	0.1	
		6.0		—	0.1	—	0.1	—	0.1	
$V_{OH}$ Output voltage High		2.0		1.9	—	1.9	—	1.9	—	
		4.5		4.4	—	4.4	—	4.4	—	
		6.0		5.9	—	5.9	—	5.9	—	
$V_{IL}$ Input voltage Low: INPI/INPB	Both standard and extended ranges	2.0	—	—	0.4	—	0.4	—	0.4	V
		4.5		—	0.9	—	0.9	—	0.9	
		6.0		—	1.2	—	1.2	—	1.2	
$V_{IH}$ Input voltage High: INPI/INPB		2.0		1.4	—	1.4	—	1.4	—	
		4.5		3.15	—	3.15	—	3.15	—	
		6.0		4.2	—	4.2	—	4.2	—	
$V_{IL}$ Input voltage Low: INPA, INPD, INPS	Both standard and extended ranges	2.0	—	—	0.3	—	0.3	—	0.3	V
		4.5		—	0.7	—	0.7	—	0.7	
		6.0		—	0.9	—	0.9	—	0.9	
$V_{IH}$ Input voltage High: INPA, INPD, INPS		2.0		1.7	—	1.7	—	1.7	—	
		4.5		3.6	—	3.6	—	3.6	—	
		6.0		4.8	—	4.8	—	4.8	—	

**CMOS GATE ARRAYS (H-Series)**

SCC0335-H, SCC0455-H  
SCC0705-H, SCC1105-H

**Preliminary**

**DC ELECTRICAL CHARACTERISTICS (Continued)**  $V_{SS} = 0V$ ; for all devices unless otherwise specified.

SYMBOL AND PARAMETER	OPERATING TEMP ( $T_{amb}$ ) <sup>1</sup>	SUPPLY VOLTAGE	TEST CONDITIONS	TEMPERATURE RANGE <sup>1</sup>						UNIT
				$T_{amb} = \text{LOW}$		$T_{amb} = 25^\circ\text{C}$		$T_{amb} = \text{HIGH}$		
				MIN	MAX	MIN	MAX	MIN	MAX	
$I_{OL}$ Output (sink) current Low driver outputs	Standard	4.5	$V_I = 0 \text{ or } 5V$ $V_{OL} = 0.26V \text{ (Std.)}$ $V_{OL} = 0.33V \text{ (Std.)}$ $V_{OL} = 0.4V \text{ (Ext.)}$	1.6	—	1.6	—	1.6	—	mA
		6.0		2.1	—	2.1	—	2.1	—	
		—		—	—	—	—	—	—	
	Extended	4.5		1.6	—	1.6	—	1.6	—	
		6.0		2.1	—	2.1	—	2.1	—	
		—		—	—	—	—	—	—	
$I_{OL}$ Output (sink) current Low buffer outputs	Standard	4.5	$V_I = 0V \text{ or } 5V$ $V_{OH} = 3.24V$ $V_{OH} = 5.34V$	3.2	—	3.2	—	3.2	—	mA
		6.0		4.2	—	4.2	—	4.2	—	
		—		—	—	—	—	—	—	
	Extended	4.5		3.2	—	3.2	—	3.2	—	
		6.0		4.2	—	4.2	—	4.2	—	
		—		—	—	—	—	—	—	
$-I_{OH}$ Output (source) current High	Standard	4.5	$V_{OH} = 3.7V$ $V_{OH} = 5.2V$	2.5	—	2.5	—	2.5	—	mA
		6.0		3.1	—	3.1	—	3.1	—	
		—		—	—	—	—	—	—	
	Extended	4.5		2.5	—	2.5	—	2.5	—	
		6.0		3.1	—	3.1	—	3.1	—	
		—		—	—	—	—	—	—	
$\pm I_{IN}$ Input leakage current	Standard	6.0	$V_{CC} \text{ or } GND$	—	—	—	0.1	—	1.0	$\mu A$
		—		—	—	—	—	—		
	Extended	6.0		—	—	—	0.1	—	1.0	
$\pm I_{OZ}$ Three-state output and open N-channel output leakage current High	Standard	5.5	$V_O = V_{CC} \text{ or } GND$	—	—	—	0.5	—	5.0	$\mu A$
		—		—	—	—	—	—		
	Extended	5.5		—	—	—	0.5	—	10.0	
		—		—	—	—	—	—	—	
$V_{TH}$ Upper threshold voltage	Standard	2.0	Internal Schmitt trigger	—	—	TBD	Typical values	—	—	V
		4.5		—	—	TBD		—	—	
		6.0		—	—	TBD		—	—	
$V_{TL}$ Lower threshold voltage		2.0		—	—	TBD	Typical values	—	—	
		4.5		—	—	TBD		—	—	
		6.0		—	—	TBD		—	—	
$V_H$ Hysteresis voltage input: INPS	—	2.0	—	—	—	TBD	Typical values	—	—	V
		4.5		—	—	TBD		—	—	
		6.0		—	—	TBD		—	—	

- NOTES:  
 1.  $T_{amb}$  Low:  $-40^\circ\text{C}$  for standard temperature range  $-55^\circ\text{C}$  for extended temperature range  $T_{amb}$  High:  $+85^\circ\text{C}$  for standard temperature range  $+125^\circ\text{C}$  for extended temperature range  
 2. Pin-connected pull-up and pull-down resistors are typically 7 to 78 K-ohms — see PERIPHERY.  
 3. When pull-up or pull-down resistors are used, current limits for  $I_{DD}$  must be extrapolated.

CUSTOM/SEMICUSTOM

Signetics

**CMOS GATE ARRAYS (H-Series)**

SCC0335-H, SCC0455-H  
SCC0705-H, SCC1105-H

**Preliminary**

**AC ELECTRICAL CHARACTERISTICS**  $V_{SS} = 0V; T_{amb} = 25^{\circ}C$

SYMBOL AND PARAMETER	PWR SUP ( $V_{DD}$ )	MIN	TYP	MAX	UNIT	SYMBOL AND PARAMETER	PWR SUP ( $V_{DD}$ )	MIN	TYP	MAX	UNIT
$f_{max}$ Maximum toggle frequency flip-flop GT00 (no set/reset)	5	—	30	—	MHz	OUTPUT STAGE TRANSITION TIMES: Input transition $\leq 20ns$ , $C_L = 50pF$ ; $V_{SS} = 0V$ , $T_{amb} = 25^{\circ}C$ .					
						$t_{THL}$ Driver outputs High-to-Low	2.0	—	130	260	ns
$f_s$ Maximum system frequency (may depend on number of gates in sequence)	5	—	12	—	MHz	$t_{THL}$ Driver outputs High-to-Low	4.5	—	26	52	ns
						$t_{THL}$ Buffer outputs High-to-Low	6.0	—	22	44	ns
$t_p$ Propagation delays for 2-input NAND gate with fanout of 2	5	—	3.5	—	ns	$t_{THL}$ Buffer outputs High-to-Low	2.0	—	65	130	ns
						$t_{THL}$ Buffer outputs High-to-Low	4.5	—	13	26	ns
						$t_{TLH}$ Buffer outputs Low-to-High	6.0	—	11	22	ns
						$t_{TLH}$ Buffer outputs Low-to-High	2.0	—	125	250	ns
$t_{TLH}$ Buffer outputs Low-to-High	4.5	—	25	50	ns	$t_{TLH}$ Buffer outputs Low-to-High	6.0	—	21	43	ns

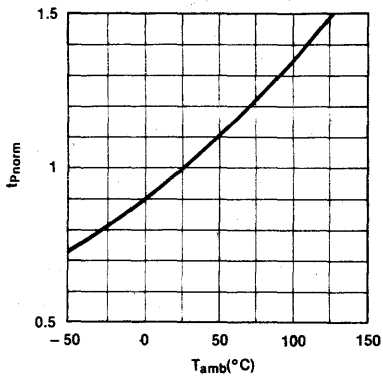


Figure 2. Normalized propagation delay ( $t_{Pnorm}$ ) as a function of the ambient temperature.

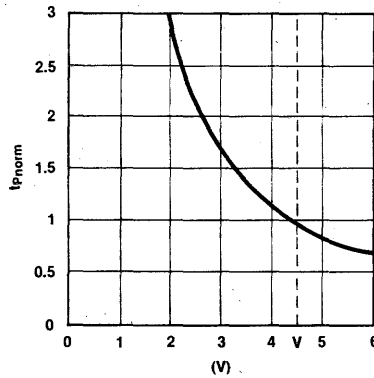


Figure 3. Normalized propagation delay ( $t_{Pnorm}$ ) as a function of the supply voltage.

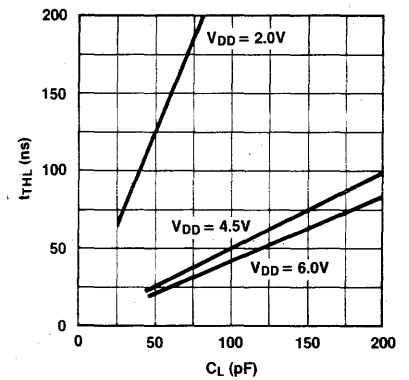


Fig. 4 Output transition time (LOW-to-HIGH) as a function of the load capacitance for driver and buffer outputs.

# CMOS GATE ARRAYS (H-Series)

SCC0335-H, SCC0455-H  
SCC0705-H, SCC1105-H

**Preliminary**

## PERIPHERY

To provide a versatile interface, H-Series CMOS arrays have numerous I/O pads—see Figure 1a. These peripheral elements can be configured to match the input or output requirements of a wide variety of logic families. Accordingly, a bonding pad may have one of the following functions assigned to it:

- **INPUT STAGE** which includes an input protection circuit (series resistor and single diode clamp to  $V_{SS}$ ). The recommended maximum load is 260 array gates, or 100 array gates for optimum speed performance. Because the input voltage is not clamped to  $V_{DD}$ , input voltages greater than the supply voltage is possible, thus allowing voltage level shifting.
- **SCHMITT TRIGGER** input stage for noise reduction, pulse shaping, or suppression of oscillation spikes associated with slow input clock transitions. The recommended maximum load is 10 array gates, or 5 for optimum speed performance.
- **TRANSCIEVER** input/output stage
- **THREE-STATE** output with driver or buffer performance capability for bussing applications
- **COMPLEMENTARY OUTPUT** with driver or buffer performance capability.
- **OPEN DRAIN** N- or P-transistor output
- **PULL-UP/PULL-DOWN** resistors (see Figure 2 for availability) may be added at various I/O stages. The values available are 5, 10, 15, 30, 60, 65, 70 and 75 Kohms.

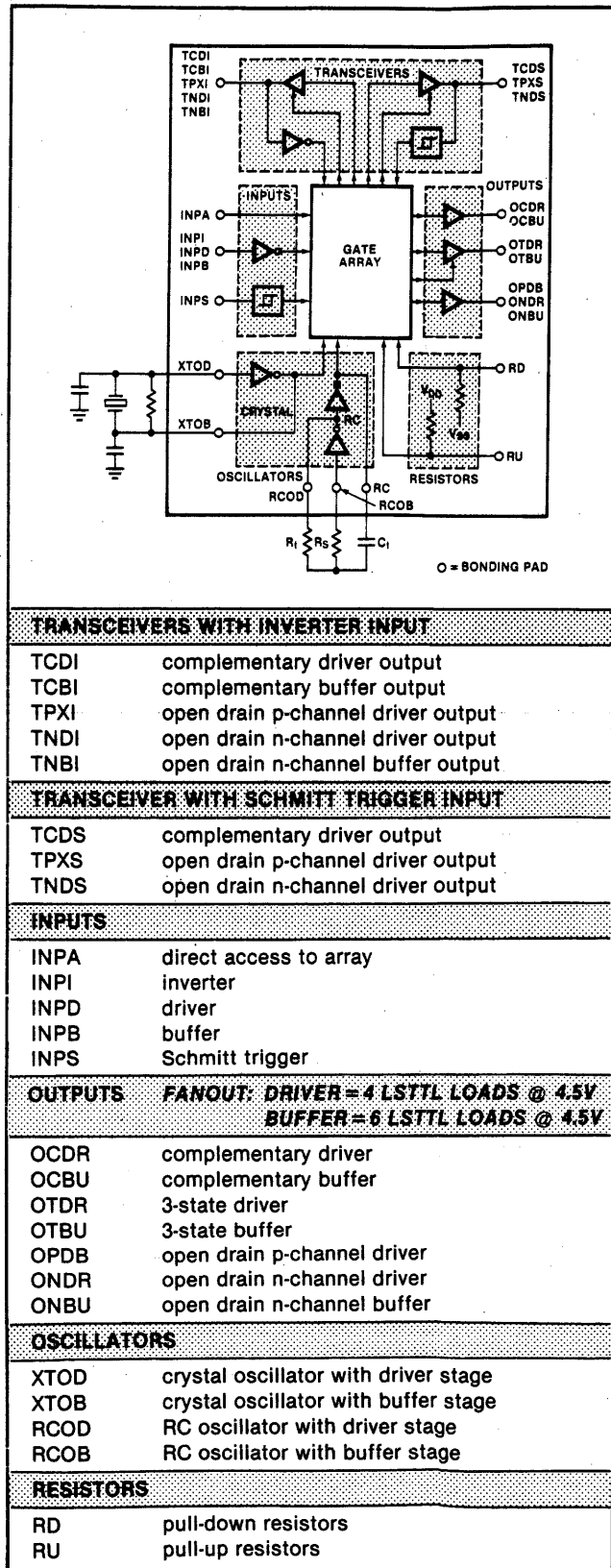


Figure 13. The SCCXXX I/O Cell Library





# Si Custom and Semicustom Circuit Capabilities

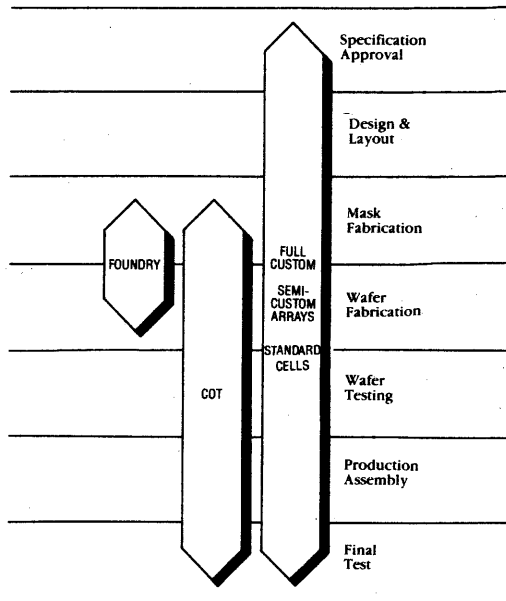
## Full custom or COT/Foundry — SSI offers it all.

SSI is committed to providing cost-effective, quality custom solutions to your most specialized applications. We offer it all: Bipolar, as well as MOS; Analog, Digital, and combinations of both.

Advanced CAD development tools enable our experienced staff of IC designers to create the optimal solution for you efficiently and accurately.

Our state-of-the-art manufacturing facility allows us to provide you with the highest quality product when you need it. If you already have your own design or tooling, SSI can provide very fast turnaround on just those foundry and full production services you require.

For low cost, quick turn around semicustom requirements, ask us about our new standard cell program. Remember, we have a program to meet all of your "application specific" integrated circuit needs.



## Semicustom Arrays

### SCA-6 AND SCA-12 SWITCHED CAPACITOR CUSTOM ARRAYS

The SCA-6 and SCA-12 are CMOS, switched-capacitor filter semicustom arrays, which can be configured to implement up to six or twelve biquadratic filter sections, other switched capacitor filter (SCF) architectures, and general analog functions. Intended primarily

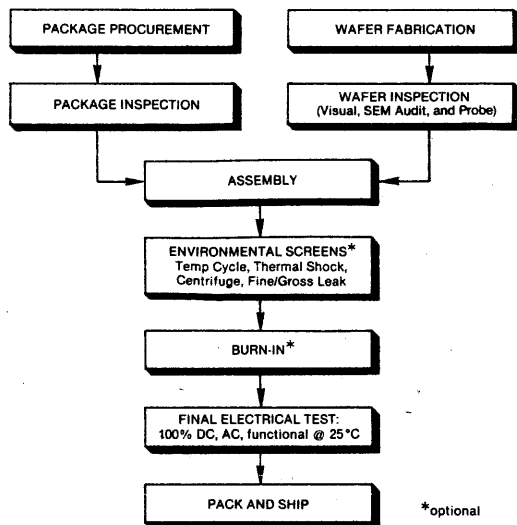
for analog signal processing applications, the SCA-6 and SCA-12 are efficient replacements for many active RC and discrete implementations, providing precision high-order filtering on a single chip.

## Quality Assurance

### SSI Quality Assurance Capabilities include:

- Environmental testing
- Burn-in
- Hi-Rel screening

### STANDARD PRODUCT Q.A. FLOW CHART



### Hi-Rel Screening Options

- Reliability Screening (Method 5004)
- Qualification and Quality Conformance Procedure (Method 5005)
- Pre-Seal Visual Inspection (Method 2010)
- Stabilization Bake (Method 1008)
- Temperature Cycle (Method 1010)
- Thermal Shock
- Constant Acceleration
- Fine Leak (Method 1014)
- Gross Leak (Method 1014)
- Burn-in
- SEM Analysis of Wafer Lots (Metallization)

### Q.A. CAPABILITIES

The quality assurance program at Silicon Systems specifies and implements standards and controls on all SSI products during all phases of design, manufacture, and assembly. Product quality is carefully monitored at each critical point. For those clients whose applications require extraordinary reliability, SSI offers special Hi-Rel screening procedures and burn-in. All Hi-Rel processing steps are in accordance with MIL Standard 883, and SSI quality assurance procedures can include any of those shown above.



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## Why Go Custom and Why Use Synertek?

Your decision to use a Custom circuit rests basically on the requirements of your system. You'll want to consider the alternatives available to you:

- **Standard circuits** are off-the-shelf products designed for general product applications. If your yearly system volume will be less than 10,000, SSI (small scale integration) and MSI (medium scale integration) circuits may be your best solution.
- **Microprocessors** may fill your need if your application requires great flexibility and cost is not an overriding factor. They are most practical when total circuit volume does not exceed 50,000.

### Advantages of Custom Circuits

A Custom circuit is an exclusive proprietary design built specifically to meet your product requirements. Its advantages are:

- **Reduced system cost** — Through circuit integration the total number of discrete and integrated components can be cut by 75-90%. This dramatically reduces component inventory, PC board assembly, and power supply costs.
- **Increased reliability** — As circuit device count and total system size is reduced, system reliability increases. For you, the more reliable a system you sell to your customers, the less you will have to expend on warranty costs.
- **Features** — Special features not available in standard circuits or microprocessors can be cost-effectively designed into a Custom circuit.
- **Market Leadership** — CUSTOM MOS/LSI technology can revolutionize a product. It enables new features to be built which would otherwise be unavailable or too expensive to implement. When your product is manufactured with a proprietary design, competitors have a more difficult time copying it. As a result, you can enjoy longer periods of market leadership and penetration.

### Why Use Synertek?

Our ability to develop sophisticated technologies insures our market leadership position in custom integrated circuits. We currently offer silicon gate HMOS, NMOS, and HCMOS technologies. Our advanced computer-aided design facilities, projection alignment equipment, 4-inch wafer fabrication lines, and VLSI testing equipment further demonstrate Synertek's commitment to state-of-the-art technology. Synertek offers the following services:

- **Classic Custom Designs** — We will design your circuit from concept. You may initially provide us with a written explanation of the function you want, or with a logic diagram of the circuit, or with just a specification.
- **Customer Design Teams (C.D.T.)** — We will train your engineers to do their own custom design. The full spectrum of design capability can be approached: classic custom hand-drawn circuits, designs using the Cell Library™, or seminars for C.O.T.™ customers. Your systems knowledge is combined with our IC design capability for optimum circuits.
- **C.O.T.™** — Synertek will become the manufacturing arm of your in-house design group or your consulting design house. By providing you with process design rules, parameters, computer simulation programs and, most of all, our total cooperation, we can assure you the manufacturing capability you want.

## Classic Custom Design

A masterpiece is never developed overnight. An artist needs time to think, plan, and create. At Synertek, the average length of time needed to bring a circuit from the concept stage to prototype production is 6-9 months. Depending on the complexity of the device, this may be longer or shorter. Simple circuits can be completed within 3-4 months. All circuits, however, are subjected to the same stringent testing, quality control, and verification checks.

Keeping with our philosophy of partnership, our engineers will confer with you often throughout the design and development phases.

### The Custom Design and Development Process

- **System definition** — Synertek and the customer establish block diagrams, flow charts, and mechanical and electrical specifications
- **Logic design and computer simulation** — Our design engineers convert system functions to MOS logic. Computer simulations of critical logic design characteristics are done in our DA (design automation) center. SCEPT™ (Synertek Circuit Emulation Program and Test) is a conventionally designed breadboard which duplicates MOS logic. Once approved by you, SCEPT™ is used to write and debug a test program. From this point, SCEPT™ is the functional reference for the remaining design steps. **1-10 weeks.**
- **Circuit design and analysis** — Individual transistors are laid out to implement the SCEPT™ logic. Particular attention is paid to critical speed paths. Additional computer-aided circuit simulation information is analyzed and incorporated into the actual circuit design. **1-4 weeks.**
- **Composite layout design** — A layout of the circuit design, called a composite, is hand drawn to minimize final chip size.
- **Digitizing** — The composite drawing is converted in our CAD (Computer Aided Design) center to a database tape using a Calma interactive graphic system. The digitized information is used to generate plots of each circuit layer. The plots are compared to the original composite and editing changes are made. Editing and checking continues until the database tape is approved for the entire composite. **1-14 weeks.**
- **Mask generation** — Once the database tape is approved a PG (pattern generation) tape is produced. This tape is used to create each mask level.  
We use three methods in mask making — photolithography, E-beam and a combination of both.
- **Prototype wafer fabrication** — During prototype fabrication, numerous quality and electrical inspections are performed to assure that every wafer lot meets our specifications.
- **First samples** — These are untested devices commonly referred to as "Cut & Go's." They are placed in ceramic packages, assembled, and sent to you for initial evaluation. **2-3 weeks.**
- **Test generation** — The Cut & Go's play an essential role in the completion of the test program, which was initiated during the circuit design stage. The test program must be verified with the Cut & Go's before it is finalized.
- **Prototype production** — After fully tested samples are approved, prototype production begins.
- **Full production** — Scheduled delivery commences after prototype qualification and test verification are completed.

## Customer Design Teams (C.D.T.)

C.D.T. is a program by which customers can be trained to do their own IC designs. It encompasses the full spectrum of custom design from the classical, hand-drawn approach, with Synertek's engineers doing the full design, to the customer owned tooling program. CDT's are designed to take advantage of the customer's system knowledge and Synertek's IC design and production capabilities. The customer has access to our facilities, design courses, the Cell Library™, advanced CAD tools, processes, design expertise, and on-going support.

The CDT design course is approximately ten weeks long and is based upon the structured design methodology. The major portion of the course demonstrates how to use the Cell Library™, with hands-on experience using our CAD tools to complete a simple design. Very little time is spent on design physics. As an on-going project we are continually developing additional short courses as updates as well as video tapes on specific topics.

Our CAD tools revolve around our mainframe computer, a VAX 11/780, to enable our customers to use commercially available software written for the VAX. NCA corporation's software is used for layout verification, including Design Rule Check, Electrical Rule Check and Network Continuity Check. The Network Continuity Check compares the layout data to the Netlist derived from schematic entry. Other services, such as sizing and PG tape generation, can also be performed on the VAX.

Silvar-Lisco's SOS package including CASS and CAL-MP is installed on the VAX. CASS is used for schematic entry via a Genisco 1000 or other graphics terminals. Once the schematic has been entered, a design data base is created from which several output data formats can be generated. The following outputs are presently available: Netlist, TEGAS for logic generation and test verification, SPICE for circuit simulation, Network Continuity Checker and CAL-MP.

CAL-MP is used for automatic placement and routing of the cells in the Cell Library™. All cells are designed to take advantage of the CAL-MP program capability. The program can handle up to 1800 cells (not gates) with 3600 cell capability to be installed soon. Chips of larger size can be created by assembling partitioned sections of up to 3600 cells each.

We are constantly updating and improving our software. Any improvements in the NCA or Silvar-Lisco software will be installed so that design capability and turn-around times can constantly be improved. Dial-up facilities to the VAX are available enabling our CDT customers to access the above software via remote terminals.

As an important part of the CDT program Synertek provides support in the form of helping with the design, instruction in the use of our CAD equipment, providing updates in the course, and providing advancements in our process capabilities. Synertek's CDT team is organized to assist the customer in all the crucial steps of the design. Once the customer reaches design proficiency, we have a separate Customer Owned Tooling (C.O.T.™) team to move designs quickly into production. As a customer's needs or designs change, CDT's are flexible enough to tailor on-going support to the specific requirements of the customer.

## The Cell Library™

The Cell Library™ evolves around our n-well, single poly, single metal 3 micron HCMOS process. The cells developed are the basic cells required to do random logic design. We will continue to design increasingly complex cells, such as multi-stage counters, voltage comparators, microprocessor cores, ALU's, etc. Customers have the option of designing and using any additional cells required in their system. These cells should conform to our format requirements, and the customer has the option of including them in the Cell Library™ user group.

Designs can be done on any graphics system with the interface preferably using GDS II formatted tapes. All cells are fully characterized over the -55°C to +125°C range. Each cell has its own data sheet, with all the electrical and topological parameters. The Cell Library™ manual is a TTL type book containing all the information on the cells.

Almost all cells are of constant height and variable width, however some cells are of different height to improve layout efficiency. For example, there is a tall and narrow flip-flop that will be used for long chains of ripple counters. Some of the basic cells, such as inverters, NOR gates, NAND gates, are of two different heights to optimize the auto-router.

Each cell has inputs at both the top and the bottom with two power lines running horizontally through the cell. The cells are self-contained, so that no external connections will be required to complete them. For most of the cells, connections are made by either poly or metal. Production type cells, such as alignment marks, CD's, etc. are also available as part of the Cell Library™.

The Cell Library™ is available to our classic custom, CDT and C.O.T.™ customers. Synertek's engineers will design a circuit for you or teach your engineers to do their own designs using the Cell Library™. C.O.T.™ customers may purchase the Cell Library™ with or without a production commitment.

Training for C.O.T.™ customers is also available for a nominal fee.

## C.O.T.™

Perhaps you have your own MOS design group, or have chosen to have your circuit designed by a consulting firm. Or maybe another MOS supplier designed the chip and you want to tool-up a second-source supplier. Whatever your design source, we can produce your circuit on a customer owned tooling (C.O.T.™) basis.

Because of our extensive experience with MOS/LSI technology, we understand your reasons for going C.O.T.™. You want to minimize design cost and production time while maximizing proprietary design control. We guarantee that your Custom circuit will receive the same confidential, proprietary treatment as our own in-house designed circuits.

You may enter the production cycle at a number of various stages. We'll accept your design on a database tape, a pattern generator tape, or working plate. You'll be given an initial documentation package that includes an overview of design rules and parameters for our MOS processes.

C.O.T.™ customers provide Synertek's Product Engineering with their chosen form of tooling along with the test tape and specifications for testing the customer's circuit.

Synertek maintains a policy of requiring characterization data for all circuits prior to transfer to production. This measure enables Synertek to do further studies on yield enhancements and correlation. Synertek strives to maximize yields at final test and

ultimately reduce circuit costs to the customer. We view C.O.T.™ as a joint effort on the part of vendor and customer. Our goal is to work with the customer's design group, to lend the necessary technical support and to build a successful working relationship.

Again, we will meet with you as early in the program as possible to establish a close working relationship. If you wish to design your own proprietary circuit, our engineering staff is available for design aid and general program guidance. We take measures to enhance a smooth product flow. Our program managers monitor your circuit from our CAD center through production. We also have a back-log control system that continually updates you on product schedules and shipments.

As a C.O.T.™ customer, you have access to our extensive manufacturing and assembly facilities in addition to our advanced processes.

### A Process for Every Masterpiece

Selecting the right process for your Custom circuit is one of your most important decisions in the design cycle.

Synertek's offering of fully proven manufacturing processes has the right answer for you. It includes state-of-the-art HMOS, HCMOS and EEPROM in addition to the industry standard NMOS silicon gate technology. Electrical and Topological Design Rules are available under a non-disclosure agreement.

## A Sample of Synertek's MOS Processes

### SILICON GATE NMOS

Process	FLDI <sup>2</sup>	ENH <sup>2</sup>	DEP <sup>2</sup>	V <sub>SB</sub>	V <sub>TEO</sub>	Volts		V <sub>TFO</sub>	V <sub>DSS</sub>	Beta A/V <sup>2</sup>	Gamma V <sup>1/2</sup>	C <sub>DA</sub> F/cm <sup>2</sup> x 10 <sup>-8</sup>	pN Ω/□	pP Ω/□	X <sub>1</sub>	Channel Length μ	Topological Pitch			Comments
						V <sub>TDO</sub>	V <sub>TFO</sub>										Poly w/s μ	Diff w/s μ	Al-Al w/s μ	
NDK4	Yes	Yes	Yes	0	+0.2	-3.0	16	10	11	0.25	0.7	15	60	1.5	6	6/6	6/6	6/6	Planox, 2 poly process, switched capacitor techniques for analog circuits	
NPM5	Yes	Yes	Yes	0	+0.5 -0.2	-3.0 -1.3	15	7	19	0.26	0.7	27	30	0.45	3 (E & D) 4 (I)	4/5	5/5	5/5	HMOSI. Dual implants for each ENH and DEP transistor (optional)*	
NPM6	Yes	Yes	Yes	0	+0.5 -0.0	-3.0 -1.3	15	7 (1.6μ)	29	0.25	0.8	24	27	0.35	2 (O) 3.5 (I)	3/3	3.5/3.5	3.5/3	Planox; single poly. Stepper technology and all dry etch process	
NFN5	Yes	Yes	Yes	0	+0.8	-2.8	27	11	17	0.75	0.7	15	65	1.2	5	5/4	5/5	5/5	Planox; E <sup>2</sup> PROM process	

\*Also available with dry etch processing at contact.

### SILICON GATE CMOS

Process	Channel	V <sub>TFO</sub>	V <sub>TO</sub>	Beta UA/V <sup>2</sup>	V <sub>DSS</sub>	Gamma	pNopP Ω/□	pPoly Ω/□	Channel Length μ	Topological Pitch				Comments
										Poly w/s	n-Well w/s	Diff. w/s	Al-Al w/s	
CPN5	P	-16	-0.9	5.5	-19	0.7	50	20	3	3/3.5	3.5/14	3/5	4/5	Single poly, n-well process
	N	+10	+0.9	17	+17	0.5	25	15	3μ	3/3.5		3/5	4/5	Available up to 12 V*
CA72	P	-16	-0.9	5.5	-19	0.7	50	1)20 15	2	2/2.5	2.5/10	2/3.5	3/3.5	Double poly, high voltage
											2/3.5	3/3.5		
	N	+10	+0.9	17	+17	0.5	25	2) <100 <100	2					

\*Also available using stepper technology.

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**ITT Multikomponent**  
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**Conseilet Promotion**  
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92150 Suresnes, France  
Tel. 5064275  
Telex (842) 614596F

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India  
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61012 Tel Aviv, Israel  
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### ITALY

**Gruppo Editoriale Jackson  
S.R.L.**  
Via Rosellini 12  
20124 Milan, Italy  
Tel. 68 80951  
Telex (843) 315366

### JAPAN

**Asahi Glass Company, Ltd.  
Electronic Components  
Group**  
1-2 Marunouchi 2-chome  
Chiyoda-ku, Tokyo 100,  
Japan  
Tel. (03) 218-5813  
Telex (781) TK4616

**Overseas Data Service Co.  
Ltd.**  
Shugetsu Building No. 12-7  
Kita-Aoyama 3-chome  
Minato-ku, Tokyo 107, Japan  
Tel. (03) 400-7090  
Telex (781) J26487

**Tokyo International  
Communications Inc.**  
Miyajima Bldg.  
28 Yoyogi 1-chome,  
Shibuya-ku  
Tokyo 151, Japan  
Tel. 379-2561  
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Box 94  
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Telex (960) 4-24143

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### SWEDEN

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Box 56,  
16126 Bromma, Sweden  
Tel. 08/252610  
Telex (854) 11181

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Taefernstrasse 15  
CH-5404 Baden-Daettwil  
Switzerland  
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Telex (845) 54070Z

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Hsin Yi Rd.  
Taipei, Taiwan, ROC  
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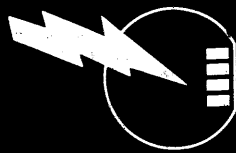
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(Mamulleri Pazarlama AS)**  
Tersane Cad. Kuthan 38/408  
Tr-Kara Koy — Istanbul,  
Turkey  
Tel. (11) 496249  
Telex 24429

### WEST GERMANY

**Astronic GmbH**  
Winzererstrasse 47d  
8000 Munich 40,  
West Germany  
Tel. (089) 309031  
Telex (841) 5216187





# Telmos Gate Arrays

## Convert Your Precision Analog or Analog-Digital Subsystem to a Single CMOS IC

Were you told to go full custom because no one could get close to your precision analog specs? Get the facts about the world's first silicon gate CMOS precision analog and analog-digital ULA—the Telmos TM 6000.

### PRECISION ANALOG SPECS

Compare the TM 6000 specs—they are 5 to 20 times more precise than a metal gate ULA.

- 90 nV/√Hz
- 5 MHz gain bandwidth product
- 0.1% capacitance matching
- 80 ppm/°C internal bandgap reference
- 5.7V buried zener (w/200Ω internal R)
- Switch cap. filter (50KHz corner frequency)
- Plus other key analog specs

### DESIGNER FLEXIBILITY VS. PREDETERMINED CELLS

Forget about rigid design rules and predetermined cells. The TM 6000 allows engineers the full use of their design creativity and imagination. It provides complete circuit flexibility and choice of designs. Send for the TM 6000 Design Manual and designers kits that consist of selected cell library configurations. All designer aids are immediately available and more are added each month.

### BUILD THESE SUBSYSTEMS AND MORE...

- Op amps (<5mV offset)
- A/D converters
- D/A converters
- Crystal or voltage controlled oscillators
- Phase locked loops
- Switched capacitor filters
- DC/DC converters
- V/F converters
- Flash A/D converters
- Analog multiplexers
- Zener/bandgap volt ref.

### AND IF ANALOG ISN'T YOUR BAG

For all digital subsystems use Telmos' high speed TM 4000 CMOS silicon gate arrays or the low cost TM 3000 metal gate arrays. For applications requiring up to 300 volt output transistors, inquire about the TM 5000 High Voltage CMOS gate arrays. Send for Design Manuals.

Call Al Kadis, *Manager of Marketing* at (408) 732-4882.

EUROPEAN OPERATIONS AND DESIGN CENTER: West Germany; Telmos GmbH, Ellingerweg 98, 8000 Munchen 80, Tel: (089) 432004.

### Send for Information/Manuals:

#### Model TM 6000 A/D Array

- Manual - \$70
- Kit Information (1-21 below)
- 1. Voltage Splitter for DC-DC converter use.
- 2. Medium performance (12 micron gate) operational amplifier.
- 3. High performance (24 micron gate) operational amplifier.
- 4. 12 micron N-channel devices with P-well contact to determine MOS characteristics. 12 micron N-channel current mirror.
- 5. 24 micron N-channel devices with P-well contact to determine MOS characteristics. 24 micron N-channel current mirror.
- 6. Second-order low-pass switched capacitor filter.

- 7. High speed comparator.
- 8. Bandgap reference.
- 9. Eight bit monotonic DAC.
- 10. High impedance N-channel and P-channel devices.
- 11. High speed differential amplifier with fixed gate.
- 12. Four bit flash converter.
- 13. 12 micron P-channel devices with substrate contact to determine MOS characteristics. 12 micron P-channel current mirror.
- 14. 24 micron P-channel devices with substrate contact to determine MOS characteristics. 24 micron P-channel current mirror.

- 15. Voltage controlled oscillator (VCO).
- 16. Triple P-channel Op-amp.
- 17. Quad P-channel Op amp.
- 18. Triple N-channel Op amp.
- 19. Quad N-channel Op amp.
- 20. Quad comparator.
- 21. Second-order high-pass switched capacitor filter.

#### Model TM 5000 High Voltage Gate

- Information

#### Model TM 4000 Silicon Gate

- Manual - \$35
- Information

#### Model TM 3000 Metal Gate

- Manual - \$35
- Information



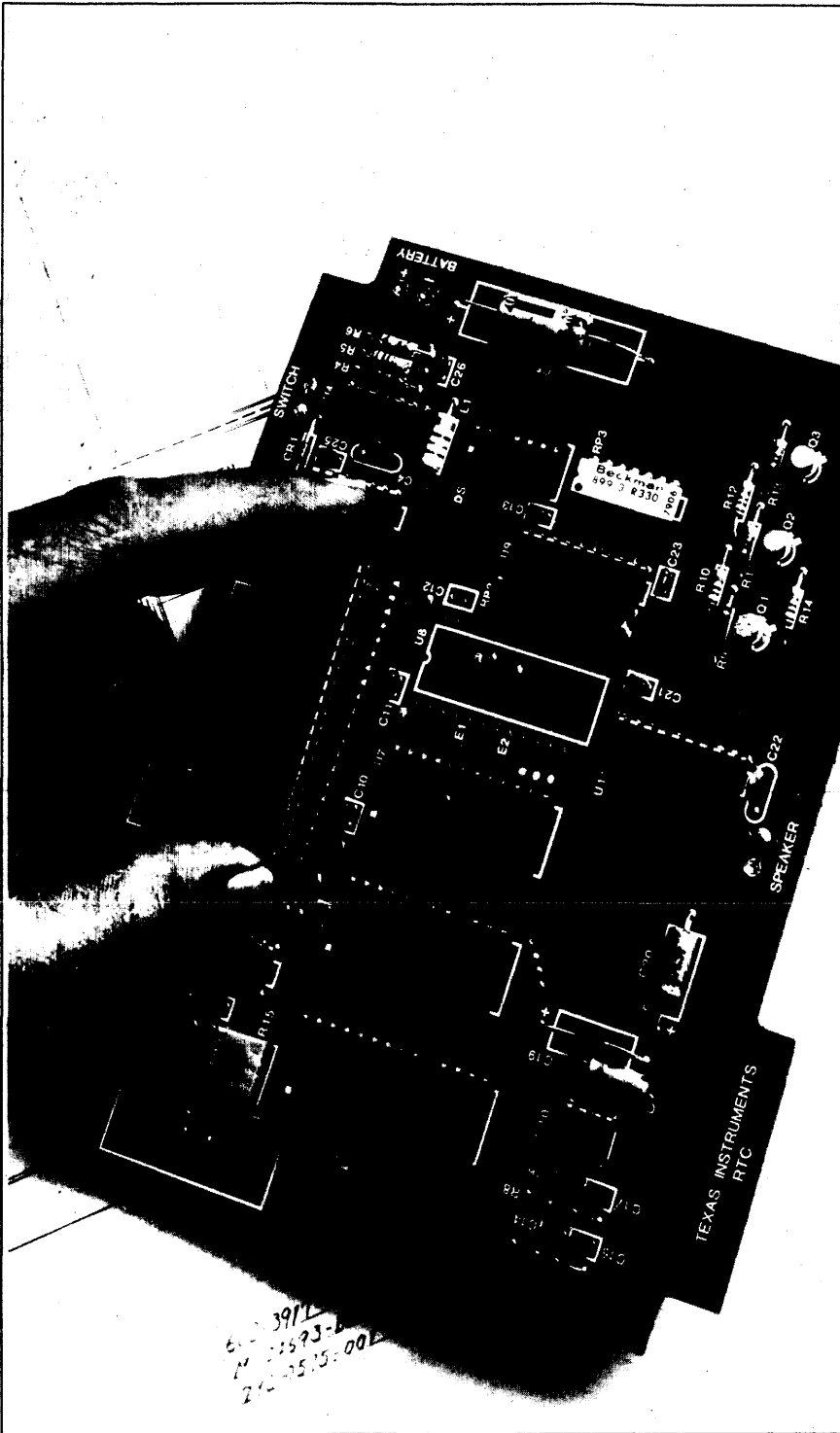
Telmos

Telmos Inc.  
740 Kifer Road  
Sunnyvale, CA 94086

# TI Standard-cell Technology:

Texas Instruments

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Now, you can achieve better performance by matching logic more closely to your design requirements. TI's new 3- $\mu$ m CMOS standard-cell technology is the fast, simple means by which you can gain the advantages of VLSI custom chips at a price you can afford. TI CMOS standard-cell technology lets you design your own semicustom ICs. Quickly. Economically. And best of all, using familiar design rules and techniques that let you focus on your logic design, rather than on IC design.

## Semicustom logic made affordable

A single semicustom chip created with TI's standard-cell technology lets you replace hundreds of discrete TTL components. Allowing you to dramatically reduce your system parts count. Improve reliability. Lower power dissipation. Develop a lighter, more compact product. And decrease assembly costs because of fewer parts and fewer inter-package connections.

What's more, system cost with TI's standard-cell logic is often much less than with discrete logic.

## Easy does it

TI's standard-cell system makes developing semicustom logic as easy as arranging packaged logic chips on a PC board. All you do is develop your design using familiar TTL logic functions as your building blocks.

Your schematic can be in the form of a discrete IC breadboard. Or you can draw your schematic on paper, or set it up using an engineering work station. Then, you create a simulation model for your schematic and exercise it to verify performance and define the tests that your chips must pass in production.

◀ Simple yet advanced, TI standard-cell technology lets you develop one-chip logic systems using traditional TTL design techniques. Extensive support from TI includes advanced engineering work stations now located at all TI Regional Technology Centers. The reduced cost, size, weight, and power consumption of TI standard-cell chips can benefit products ranging from simple electronic games to logic-intensive systems for advanced computers.

## PRODUCT PREVIEW

This page contains information on a product under development. Texas Instruments reserves the right to change or discontinue this product without notice.

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# Semicustom CMOS logic you design using your TTL know-how.

## TYPICAL STANDARD-CELL DESIGN

### APPLICATION

- FLOPPY-DISK LOGIC DISPLACEMENT—8-BIT MICROPROCESSOR DESIGN WITH ROM AND DRAM INTERFACES
- PRESENT DESIGN—48 LPS AND SCHOTTKY DEVICES
- THIS DESIGN—1-40 PIN 1,550-GATE STANDARD CELL

### PERFORMANCE ANALYSIS

#### IMPACT AREA

	LPS DESIGN	STANDARD CELL
PERFORMANCE	LPS	LPS EQUIVALENT OR BETTER
POWER DISSIPATION	1.35 W	50 mW
PC BOARD AREA	40 SQ. INCHES	3 SQ. INCHES
NO. OF SOLDER JOINTS	720	40
COMPONENT COST	X	0.3X

#### DESIGN DELIVERY TIMING

#### ACTIVITY

	WEEKS	CUM.	COMMENTS
DESIGN IDENTIFIED	0	0	QUOTE, TERMS, AND CONDITIONS
DESIGN SUBMITTED TO RTC	4	4	LPS SCHEMATICS AND TEST DATA
RTC SUBMISSION TO FACTORY	6	10	COMPUTER-GENERATED DATA BASE AND LAYOUT
FACTORY SUPPLIES PROTOTYPES	7	17	20 UNITS TO SPECIFICATION
FIRST PRODUCTION DELIVERY	6	23	HIGH-VOLUME MANUFACTURE

Low logic costs per installed gate combine with quick turnaround times to help bring your system cost down the learning curve faster.

After that, your responsibilities are over. TI handles all the time-consuming details of chip layout, testing, and verification. Turnaround time and costs are kept to a minimum by TI's automated design system.

This approach gives you two key advantages. First, it lets you capitalize on the expertise you've developed over years of using standard logic families like SN54/74LS Low-power Schottky TTL. Second, it gives you the benefit of TI's decades of experience in IC design.

If you prefer, you can do the chip layout yourself. In this case, TI supplies the specifications you need to lay out your design using standard cells.

### Wide choice of functions eases design task

Many popular functions equivalent to SN54/74 devices are already available in TI's new SN54/74SC family of standard-cell logic functions. The current family members are 3- $\mu$ m twin-well CMOS parts based on TI's SN54/74HC family of high-speed CMOS logic functions. The new SN54/74SC family meets or exceeds SN54/74HC specifications and offers the same advantages of low power consumption, high noise immunity, flexible power-supply requirements, and switching speeds comparable to Low-power Schottky performance.

With a 2-picofarad load, the typical propagation delay is 3 ns per gate, with a typical power dissipation of 60  $\mu$ W per gate at 1 MHz. Electrostatic discharge protection on input and output buffers exceeds 2000 volts. Latchup protection

on outputs has been shown to exceed  $\pm 300$  mA at room temperature.

Many new standard-cell functions are coming soon. Plans include unique "procedural cells" that let you partially define your own on-chip RAM, ROM, ALU, counters, and other functions. Also upcoming is a series of advanced standard-cell functions offering Advanced Low-power Schottky performance levels.

Depending on your design needs, TI standard-cell ICs can be packaged in anything from 8-pin DIPs to multipin chip carriers. And MIL versions will also be available. †

### Sized to fit

What's more, standard-cell ICs are sized to meet your specific design requirements. You aren't tied to a specific chip size, as in the case of gate or logic arrays.

†Check field sales office for availability.

Standard-cell functions are also packed much more densely than logic-array functions. That's because the functions are separately selected and placed on chip for each design. In contrast, logic arrays must implement the desired function: using a preset collection of logic elements, such as NOR gates.

TI standard-cell ICs let you avoid the problem of "unused" gates. They also minimize silicon-consuming routing areas needed to interconnect logic arrays.

The result is cost-effective, compact ICs that pack much more functionality in less chip space.

This translates into substantial savings in IC production costs. Savings that make TI standard-cell solutions an attractive choice for low- as well as high-volume products.

### Versatile logic ICs

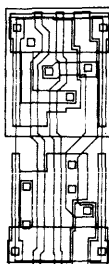
You can put standard-cell ICs from TI to work in just about any system requiring digital logic. Like printers and terminals. Dedicated controllers. Instrumentation. Automotive controls. Video games, personal computers, and other consumer electronic products.

Standard-cell ICs from TI provide the affordable logic you need to build better man-to-machine and machine-to-machine interfaces into your products. For "friendlier" computers and "smarter" office systems.

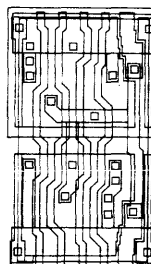
The modest power, size, and weight requirements of TI's CMOS standard-cell chips are also ideal for battery-powered and portable equipment.

## TYPICAL CELLS

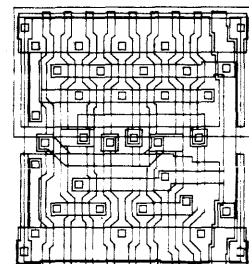
MINIMUM SIZE  
LOW DRIVE



MEDIUM SIZE  
MODERATE DRIVE



MAXIMUM SIZE  
HIGH DRIVE



Each function in the SN54/74SC family has a number of physical implementations, allowing TI to better optimize your circuit design.

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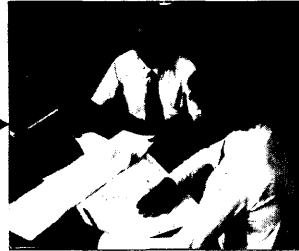
# Simple design system with quick

## Texas Instruments step-by-step approach speeds development of SN54/74SC semicustom logic.

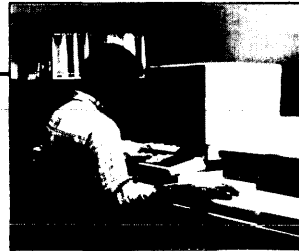
TI's standard-cell integration system offers two basic routes for developing standard-cell ICs. With one, you design the logic system you need and leave the chip layout and verification to TI. Or you can do the layout yourself, using standard-cell specifications. Here's the typical design sequence you'll follow if TI does the layout:



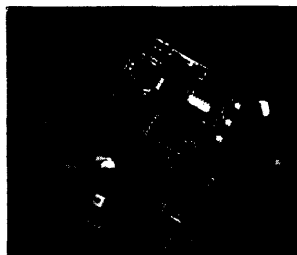
**1** You develop a schematic or discrete IC breadboard of the required logic system using SN54/74SC functions (your schematic can be drawn on paper or via an engineering work station).



**2** You create a TIDAL\* data base for your design using a schematic-capture program. This can be done either at your plant site using an engineering work station or at your local TI Regional Technology Center.



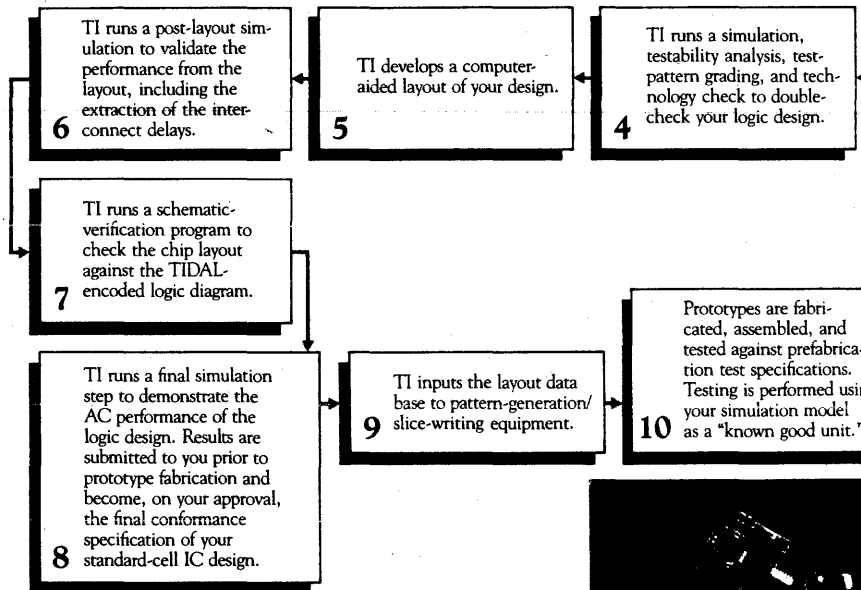
**3** You input your TIDAL data base to various design tools to perform simulation, testability analysis, test generation, and test-pattern grading. All these tools are available through your local TI Technology Center, along with computers to run them. Some of the tools can be run on your own computer or accessed via a remote-job-entry connection (depending on the type of computer you have).



**10** Prototypes are fabricated, assembled, and tested against prefabrication test specifications. Testing is performed using your simulation model as a "known good unit."

**11** Prototypes are shipped to you for in-system evaluation and the decision to go to production.

\*TI's high-level hardware description language. Trademark of Texas Instruments.



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# turnaround.

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## All the support you need... nearby

The support you need to create your standard-cell design is conveniently available at your nearest TI Regional Technology Center.

Each Technology Center is staffed with a team of standard-cell specialists who will work closely with you to coordinate TI's layout, testing, and fabrication efforts with your design goals and product schedule.

Support services include a telephone hot line for getting fast answers to technical questions. In-depth training courses on standard-cell technology. Assistance in design optimization and test development. Access to hardware and software design tools. Even complete turnkey design services. With the all-important personal contact that ensures a quick response to your needs.

## Streamlined design system

TI's standard-cell integration system features a set of state-of-the-art, fully integrated design-automation tools. These range from schematic-capture and design-simulation programs to design-verification software and automated tooling equipment.

All the tools can share design information via TI's TIDAL Transportable Integrated Design Automation Language. This saves valuable engineering time by eliminating the need to recode incompatible programs.

Many of the design tools can run on your own system (depending on the level of the tool). Systems like the IBM personal computer, IBM 4341, VAX 11/780™ systems, and a wide variety of engineering work stations.

Some of the software tools are also available for licensing. Dial-up access to the more complex design software is available through the Regional Technology Centers. In addition, engineering work stations are located at each of the Technology Centers for designers who wish to come in and use them.

TI's design system is the outcome of two decades of involvement in design

automation. Many of the tools—such as TI's proprietary layout-verification software—have provided major productivity improvements on scores of TI designs for ICs. This sophisticated design system gives your standard-cell design a very high probability of first-pass prototypes which are not only functional but producible.

## Fast turnaround

TI's design system cuts manual intervention in the design process to a minimum, resulting in very short design-cycle times.

To speed things up still further, TI's Customized Components Division (the division responsible for TI's standard-cell product area) has its own dedicated wafer-fabrication facility and assembly

and test facility for prototype and low-volume IC production.

In most cases, you'll have prototypes in your hands within six to eight weeks after joint approval of design data base. ‡ Our objective is to be fast—our commitment is to provide on-time delivery of ICs that fully meet your specifications. Providing the crucial edge you need in getting a product to market ahead of your competition.

## For more information

To find out more about the TI standard-cell integration system, please contact your nearest TI field sales office as listed on the back cover.

‡Cycle time from initial design to prototype shipments is typically 10 to 12 weeks.

## Texas Instruments SN54/74SC Functions

- GATES**
- 74SC00 2-INPUT NAND
  - 74SC02 2-INPUT NOR
  - 74SC04 INVERTER
  - 74SC08 2-INPUT AND
  - 74SC10 3-INPUT NAND
  - 74SC11 3-INPUT NAND
  - 74SC20 4-INPUT NAND
  - 74SC21 4-INPUT AND
  - 74SC22 5-INPUT NAND
  - 74SC27 3-INPUT NOR
  - 74SC30 8-INPUT NAND
  - 74SC32 2-INPUT OR
  - 74SC86 2-INPUT XOR
  - 74SC260 5-INPUT NOR
  - 74SC266 2-INPUT XOR
  - 74SC425 3-STATE BUFFER, LOW ENABLE
  - 74SC426 3-STATE BUFFER, HIGH ENABLE
  - 74SC2000 6-INPUT NAND
  - 74SC2001 7-INPUT NAND
  - 74SC2002 6-INPUT OR
  - 74SC2003 7-INPUT OR
  - 74SC2004 5-INPUT AND
  - 74SC2005 6-INPUT AND
  - 74SC2006 7-INPUT AND
  - 74SC2007 8-INPUT AND
  - 74SC2008 5-INPUT OR
  - 74SC2009 6-INPUT OR
  - 74SC2010 7-INPUT OR
  - 74SC2011 8-INPUT OR
  - 74SC4002 4-INPUT NOR
  - 74SC4072 4-INPUT OR
  - 74SC4075 3-INPUT OR
  - 74SC4078 8-INPUT NOR
- FLIP-FLOPS, REGISTERS, LATCHES**
- 74SC74 D-TYPE FF
  - 74SC75 TRANSPARENT LATCH
  - 74SC109 J-K BAR FF, PET
  - 74SC112 J-K FF, NET
  - 74SC373 OCTAL TRANSPARENT LATCH
  - 74SC374 OCTAL D-TYPE REGISTER
  - 74SC2101 D-TYPE FF WITH PARALLEL LOAD
  - 74SC2102 T-TYPE FF
  - 74SC2103 T-TYPE FF WITH PARALLEL LOAD
  - 74SC2104 GATED T-TYPE FF
  - 74SC2105 GATED T-TYPE FF WITH PARALLEL LOAD
  - 74SC2106 GATED T-BAR FF
  - 74SC2107 GATED T-BAR FF WITH PARALLEL LOAD

- COUNTERS, TIMERS**
- 74SC192 UNIVERSAL 4-BIT DECADE COUNTER
  - 74SC193 UNIVERSAL 4-BIT BINARY COUNTER
  - 74SC194 UNIVERSAL 4-BIT SHIFT REGISTER
  - 74SC198 UNIVERSAL 8-BIT SHIFT REGISTER
  - 74SC2200 UNIVERSAL 8-BIT BINARY COUNTER
  - 74SC2201 UNIVERSAL 8-BIT DECADE COUNTER
- DECODERS, MULTIPLEXERS**
- 74SC138 3:8 DECODER
  - 74SC139 2:4 DECODER
  - 74SC150 16:1 MUX
  - 74SC154 4:16 DECODER
  - 74SC251 8:1 MUX
  - 74SC253 4:1 MUX
  - 74SC258 2:1 MUX
- COMPARATORS, PARITY GENERATORS**
- 74SC85 4-BIT COMPARATOR
  - 74SC280 PARITY GENERATOR
  - 74SC348 PRIORITY ENCODER
  - 74SC688 8-BIT COMPARATOR
- ADDERS, ALUs, SUPPORT FUNCTIONS**
- 74SC83 4-BIT ADDER
  - 74SC181 ALU
  - 74SC182 CARRY GENERATOR
  - 74SC482 CONTROL ELEMENT
  - 74SC2300 RC CRYSTAL OSCILLATOR
- ROM, RAM**
- 74SC189 16 × 4 RAM
  - 74SC371 256 × 8 ROM
- INPUT BUFFERS**
- 74SC5000 CMOS INPUT BUFFER
  - 74SC5001 TTL INPUT BUFFER
  - 74SC5002 2 mA BIDIRECTIONAL I/O BUFFER
  - 74SC5003 SCHMITT INPUT BUFFER
  - 74SC5004 4 mA BIDIRECTIONAL I/O BUFFER
- OUTPUT BUFFERS**
- 74SC5100 2 mA OUTPUT BUFFER
  - 74SC5101 2 mA OUTPUT BUFFER, 3-STATE
  - 74SC5102 2 mA OUTPUT BUFFER, OPEN DRAIN
  - 74SC5103 4 mA OUTPUT BUFFER
  - 74SC5104 4 mA OUTPUT BUFFER, 3-STATE
- Planned availability 4Q83 ● Planned availability 1H84

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# TEXAS INSTRUMENTS

## SN54/74SC STANDARD CELL

### PRODUCT SUMMARY

FUNCTION	DESCRIPTION	C <sub>i</sub> (pF)	DELAY t <sub>pD</sub> (ns)†	dDELAY d (t <sub>pD</sub> ) (ns/pF)†
<b>Internal Cell Functions:</b>				
<b>Inverters</b>				
SN54/74SC04	Always Active			
IV110	1X (Minimum Drive Capability .....	0.15	2.8	1.5
IV120	2X .....	0.32	1.8	0.8
IV140	4X .....	0.67	1.2	0.5
IV160	6X .....	1.01	1.0	0.4
IV180	8X .....	1.34	0.9	0.3
<b>NAND Gates</b>				
SN54/74SC00	2-Input NAND			
NA210	1X .....	0.15	3.3	1.9
NA220	2X .....	0.33	2.0	0.9
NA230	3X .....	0.50	1.8	0.6
NA240	4X .....	0.68	1.5	0.5
NA260	6X .....	1.07	1.4	0.4
SN54/74SC10	3-Input NAND			
NA310	1X .....	0.12	4.2	2.5
NA320	2X .....	0.31	2.5	1.1
NA330	3X .....	0.48	2.2	0.7
NA340	4X .....	0.65	2.0	0.5
SN54/74SC20	4-Input NAND			
NA410	1X .....	0.14	5.0	3.0
NA420	2X .....	0.32	3.2	1.2
NA430	3X .....	0.48	2.8	0.8
<b>NOR Gates</b>				
SN54/74SC02	2-Input NOR			
NO210	1X .....	0.12	4.3	2.6
NO220	2X .....	0.29	2.5	1.3
NO230	3X .....	0.43	2.1	0.8
NO240	4X .....	0.58	1.8	0.6
SN54/74SC27	3-Input NOR			
NO310	1X .....	0.10	5.9	3.8
NO320	2X .....	0.27	3.7	1.9
NO330	3X .....	0.41	3.0	1.2
SN54/74SC4002	4-Input NOR			
NO410	1X .....	0.13	7.5	4.3
NO420	2X .....	0.24	4.9	2.3

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FUNCTION	DESCRIPTION	C <sub>i</sub> (pF)	DELAY t <sub>pD</sub> (ns) †	dDELAY d (t <sub>pD</sub> ) (ns/pF) †
<b>AND Gates</b>				
SN54/74SC08	2-Input AND			
AN210	1X .....	0.15	3.8	1.0
AN220	2X .....	0.15	3.5	0.5
AN240	4X .....	0.15	3.1	0.3
AN260	6X .....	0.33	2.8	0.2
SN54/74SC11	3-Input AND			
AN310	1X .....	0.14	4.5	1.2
AN320	2X .....	0.14	4.1	0.7
AN340	4X .....	0.14	4.6	0.5
AN360	6X .....	0.32	3.5	0.3
<b>OR Gates</b>				
SN54/74SC32	2-Input OR			
OR210	1X .....	0.17	4.3	1.4
OR220	2X .....	0.17	4.0	0.8
OR240	4X .....	0.29	3.3	0.4
OR260	6X .....	0.29	3.7	0.3
SN54/74SC4075	3-Input OR			
OR310	1X .....	0.14	5.2	1.4
OR320	2X .....	0.14	5.0	0.9
OR340	4X .....	0.24	4.3	0.5
OR360	6X .....	0.24	5.1	0.4
SN54/74SC4072	4-Input OR			
OR410	1X .....	0.12	6.5	1.5
OR420	2X .....	0.12	6.2	0.9
OR440	4X .....	0.23	5.5	0.5
OR460	6X .....	0.23	5.8	0.4
<b>FLIP-FLOPS</b>				
SN54/74SC74	D-Type FLIP-FLOPS			
DFB20	Preset & Clear .....	0.13	8.8	0.7
DFP20	Preset .....	0.13	7.9	0.7
DFC20	Clear .....	0.13	7.9	0.7
DFN20	No Preset or Clear .....	0.16	7.0	0.7
DFY20	D Low, Preset .....	0.29	7.3	0.7
DFZ20	D Low, Preset & Clear .....	0.29	7.9	0.7
<b>External Input and Output (I/O) Cell Functions:</b>				
<b>Input Cells</b>				
SN54/74SC5000	CMOS Level Input, Inverting			
IPA00	Minimum Height Cell .....	1.0	1.6	0.7
IPB00	Corner Cell .....	1.0	1.6	0.7
IPC00	Minimum Width Cell .....	1.0	1.6	0.7
SN54/74SC5001	TTL Level Input, Inverting			
IPA03	Minimum Height Cell .....	1.7	1.6	0.4
IPB03	Corner Cell .....	1.7	1.6	0.4
IPC03	Minimum Width Cell .....	1.7	1.6	0.4

**PRODUCT PREVIEW**

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INSTRUMENTS**

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FUNCTION	DESCRIPTION	C <sub>i</sub> (pF)	DELAY t <sub>pD</sub> (ns) <sup>†</sup>	dDELAY d (t <sub>pD</sub> ) (ns/pF) <sup>†</sup>
<b>Output Cells</b>				
SN54/74SC5100	Push-Pull, 2 mA, Non-Inverting			
OPA20	Minimum Height Cell .....	0.53	8.6	—
OPB20	Corner Cell .....	0.53	8.6	—
OPC20	Minimum Width Cell .....	0.53	8.6	—
SN/74SC5101	3-State, 2 mA, Non-Inverting			
OPA22	Minimum Height Cell .....	0.53	12.0	—
OPB22	Corner Cell .....	0.53	12.0	—
OPC22	Minimum Width Cell .....	0.53	12.0	—
SN54/74SC5102	Open-Drain, 2 mA, Non-Inverting			
OPA21	Minimum Height Cell .....	0.56	8.4	—
OPB21	Corner Cell .....	0.56	8.4	—
OPC21	Minimum Width Cell .....	0.56	8.4	—

<sup>†</sup> Delays are nominal values at room temperature and 5.0 volts. Internal cell delays are at 1.0 pF load. Output cell delays are at 15 pF. Gate delays and delta delays are average of t<sub>pHL</sub> and t<sub>pLH</sub>. Flip-flop delays are average of clock to "Q" and clock to "Q-BAR". These delays are presented to illustrate the typical capability of the SN54/74SC CMOS standard cell family. Please consult the appropriate data sheet before starting design.

**PRODUCT PREVIEW**

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# TEXAS INSTRUMENTS

## SN54/74SC STANDARD CELL

### PLANNED PRODUCTS

FUNCTION	DESCRIPTION
<b>Internal Gates</b>	
● 54/74SC425	Buffer, 3-State, Active Low
● 54/74SC426	Buffer, 3-State, Active High
▲ 54/74SC240	Inverter, 3-State, Active Low
▲ 54/74SC231	Inverter, 3-State, Active High
● 54/74SC11	3-Input AND
● 54/74SC21	4-Input AND
▲ 54/74SC2004	5-Input AND
▲ 54/74SC2005	6-Input AND
▲ 54/74SC2006	7-Input AND
▲ 54/74SC2007	8-Input AND
● 54/74SC22	5-Input NAND
▲ 54/74SC2000	6-Input NAND
▲ 54/74SC2001	7-Input NAND
● 54/74SC30	8-Input NAND
▲ 54/74SC2008	5-Input OR
▲ 54/74SC2009	6-Input OR
▲ 54/74SC2010	7-Input OR
▲ 54/74SC2011	8-Input OR
● 54/74SC260	5-Input NOR
▲ 54/74SC2002	6-Input NOR
▲ 54/74SC2003	7-Input NOR
● 54/74SC4078	8-Input NOR
● 54/74SC86	Exclusive OR
● 54/74SC266	Exclusive NOR
<b>Decoders/Demultiplexers</b>	
● 54/74SC139	2:4 Decoder/Demultiplexer
● 54/74SC138	3:8 Decoder/Demultiplexer
● 54/74SC154	4:16 Decoder/Demultiplexer
<b>Multiplexers</b>	
● 54/74SC258	2:1 MUX, 3-State, Enable
● 54/74SC153	4:1 MUX, With or Without Enable
● 54/74SC253	4:1 MUX, 3-State, Enable
● 54/74SC152	8:1 MUX, With or Without Enable
● 54/74SC251	8:1 MUX, 3-State, Enable
▲ 54/74SC150	16:1 MUX, Enable

FUNCTION	DESCRIPTION
<b>Bistable Latches</b>	
● 54/74SC75	Active High Enable, or Set/Reset
● 54/74SC373	D-Type Latch, 3-State
<b>Flip-Flops</b>	
● 54/74SC109	J-K, Positive Edge Triggered
● 54/74SC112	J-K, Negative Edge Triggered
▲ 54/74SC374	D-FF, 3-State
● 54/74SC2101	D-FF, ASYNC Load
● 54/74SC2102	Toggle-FF
● 54/74SC2103	Toggle-FF, ASYNC Load
▲ 54/74SC2104	T-FF, Active High
▲ 54/74SC2105	T-FF, Active High, ASYNC Load
▲ 54/74SC2106	T-FF, Active Low
▲ 54/74SC2107	T-FF, Active Low, ASYNC Load
<b>Input, Output &amp; I/O Cells</b>	
● 54/74SC5003	Input Buffers with Hysteresis
● 54/74SC5002	Bidirectional Buffers, 2 mA OUT
▲ 54/74SC5004	Bidirectional Buffers, 4 mA OUT
▲ 54/74SC5005	Bidirectional Buffers, 6 mA OUT
▲ 54/74SC5103	Output Buffers, 4 mA OUT
▲ 54/74SC5105	Output Buffers, 6 mA OUT
▲ 54/74SC5104	3-State Output Buffers, 4 mA OUT
▲ 54/74SC5106	3-State Output Buffers, 6 mA OUT
<b>Counters &amp; Shift Registers</b>	
▲ 54/74SC192	Universal 4-Bit Decade Counter
▲ 54/74SC193	Universal 4-Bit Binary Counter
▲ 54/74SC2200	Universal 8-Bit Binary Counter
▲ 54/74SC2201	Universal 8-Bit Decade Counter
▲ 54/74SC194	Universal 4-Bit Shift Register
▲ 54/74SC198	Universal 8-Bit Shift Register

**Planned Release Dates**

- 4Q83    ▲ 1H84    △ 2H84

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FUNCTION	DESCRIPTION
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### Comparators, Parity Generators

- ▲ 54/74SC85 4-Bit Comparator
- ▲ 54/74SC688 8-Bit Comparator
- ▲ 54/74SC280 Parity Generator
- ▲ 54/74SC348 8:3 Priority Encoder, 3-State

### Adders, Alus, Support Functions

- ▲ 54/74SC83 4-Bit Adder
- ▲ 54/74SC181 ALU
- ▲ 54/74SC182 Carry Generator
- 54/74SC2300 Crystal Oscillator

### RAM, ROM, PLA

- 54/74SC189 16 × 4 RAM
- ▲ 54/74SC371 256 × 8 ROM
- 54/74SC2500 PLA

FUNCTION	DESCRIPTION
----------	-------------

### Procedural Cells

- 54/74SC3000 RAM, 1K, 2-256 Words × 4-32 Bits
- △ 54/74SC3012 RAM, Dual Port
- △ 54/74SC3013 RAM, Serial Access
- ▲ 54/74SC3001 ROM
- 54/74SC3002 PLA
- ▲ 54/74SC3003 Universal Shift Register
- ▲ 54/74SC3004 Synchronous Binary Counter
- ▲ 54/74SC3005 Synchronous Decimal Counter
- ▲ 54/74SC3006 Register File
- △ 54/74SC3014 Register File with Stack
- △ 54/74SC3015 Register File, Dual Port
- 54/74SC3007 Arithmetic Logic Unit
- 54/74SC3008 Synchronous Full Adder
- ▲ 54/74SC3009 Comparator, Magnitude
- ▲ 54/74SC3010 Comparator, Equals
- △ 54/74SC3016 Multiplier
- △ 54/74SC3017 Barrel Shifter
- ▲ 54/74SC3011 LCD Driver

### Planned Release Dates

- 4Q83
- ▲ 1H84
- △ 2H84

TEXAS  
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▲ 54/74SC198	Universal 8-Bit Shift Register

Planned Release Dates:

● 4Q83    ▲ 1H84    △ 2H84

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FUNCTION	DESCRIPTION
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- △ 54/74SC3016 Multiplier
- △ 54/74SC3017 Barrel Shifter
- ▲ 54/74SC3011 LCD Driver

# LOW COST ARRAYS FOR TTL REPLACEMENT

# TAL-002, TAL-004 LOW POWER SCHOTTKY LOGIC ARRAYS

JUNE 1983 – REVISED OCTOBER 1983

- Family

TWO ARRAYS	INTERNAL GATES	I/O BUFFERS
TAL-002	320	28
TAL-004	500	36

- Technology

Established Double Level Metal Low Power Schottky

- Layout

Designed for the user at His Own Location or by TI Regional Technology Centers

- Internal Gates

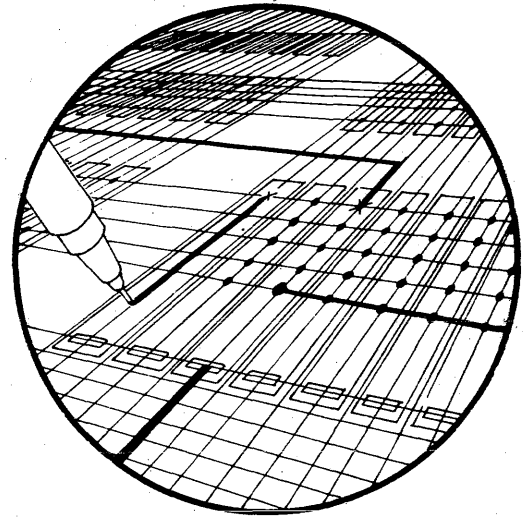
- Single 5-V Power Supply
- 5 ns Maximum Gate Delay at 1.25 mW Gate Dissipation
- Toggle Rate 25 MHz
- Twenty-Two Standard Macros currently available

- I/O Buffers

- Standard True and Complement Input
- NAND Output
- Three-State Output
- Standard I/O
- Open-Collector NAND Output
- Schmitt Trigger Input
- Power-Up Reset Circuit

- Design Verification

- Closed Loop Logic to Layout Check
- Test Pattern and Design Rule Check
- Logic, Timing and Fault Simulation
- Easy Access by User to TI Software



## description

The TAL family of devices are Low Power Schottky master slices which are held in stock at TI. The master slices are converted into LSI slices by the deposition of two layers of metal according to an interconnection pattern drawn by the user on a TI supplied translucent grid. The resulting devices are tested according to input-output logic patterns devised by the user, and become the user's proprietary LSI parts.

A comprehensive design manual sets out the logic and layout rules for the process and also explains a simple shorthand method of describing the circuit so that a detailed check of layout and test patterns can be carried out at TI. This computer aided check comprehends logic, layout, and test patterns and results in a dialogue in which the user and TI work together to eliminate errors prior to mask fabrication.

The layout task is considerably eased by the availability of standard overlays for commonly used logic functions and standard buffers are provided for inputs and outputs. The user is also able to specify his own blocks of logic in cases where a number of identical pieces of circuit occur in the design.

As can be seen from Figure 1, the array consists of blocks of five input NAND gates of which any four are useable in any one cell. The gates consist of output transistor T1 and multi-emitter transistor T2. The emitter of T1 is normally connected to ground and its collector can feed up to eight subsequent multi-emitter inputs within the array. The logic design rules for the interconnection of gates are explained in detail in the design manual.

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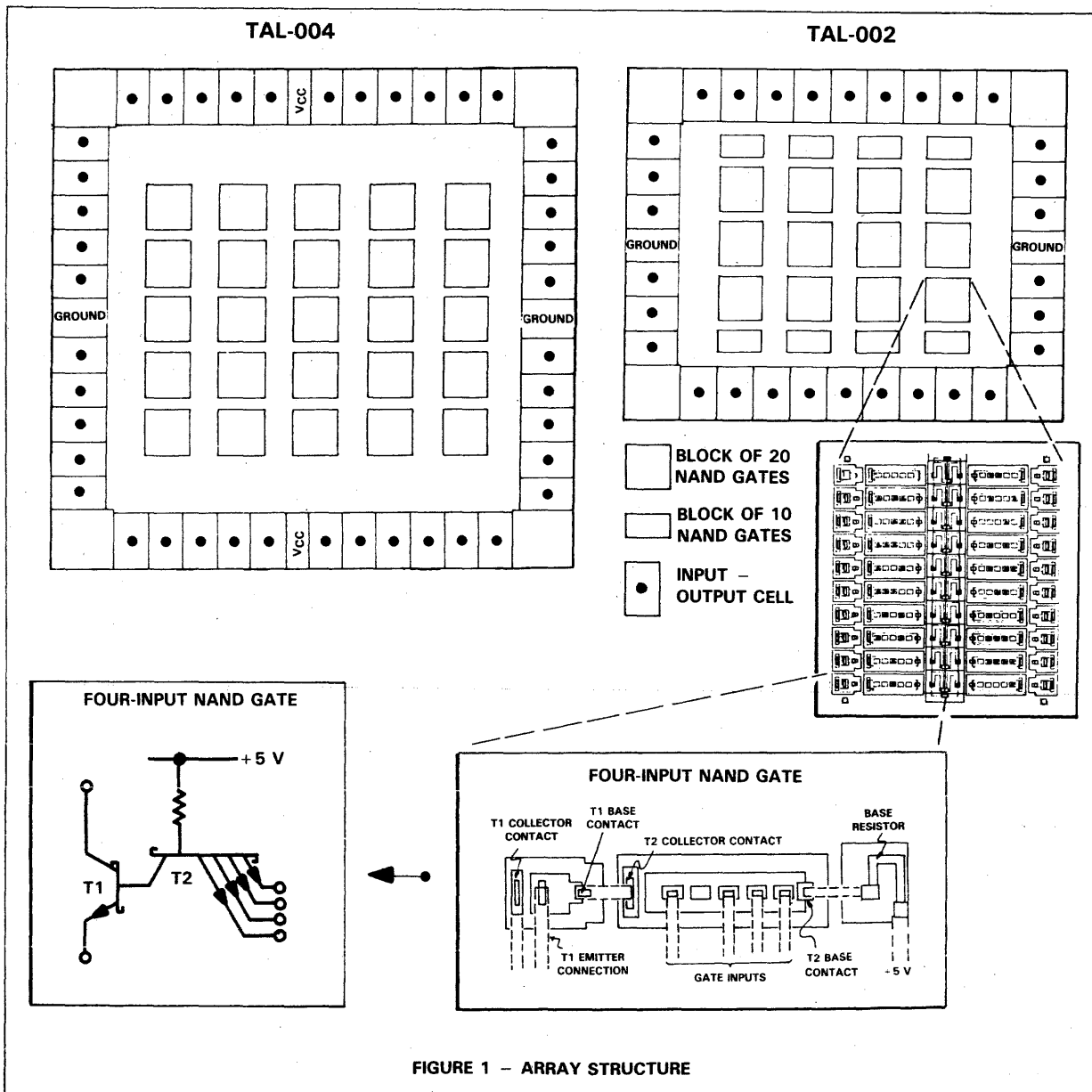
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**TAL-002, TAL-004  
LOW POWER SCHOTTKY  
LOGIC ARRAYS**

Texas Instruments

CUSTOM/SEMICUSTOM



**FIGURE 1 - ARRAY STRUCTURE**

The TAL-004 has 500 internal four-input NAND gates and the TAL-002 has 320. A single 5-V power supply is needed for the arrays and both device types are available in 28- and 40-pin DIL plastic packages, with the option also, of a 20-pin package for the TAL-002.

**manual layout**

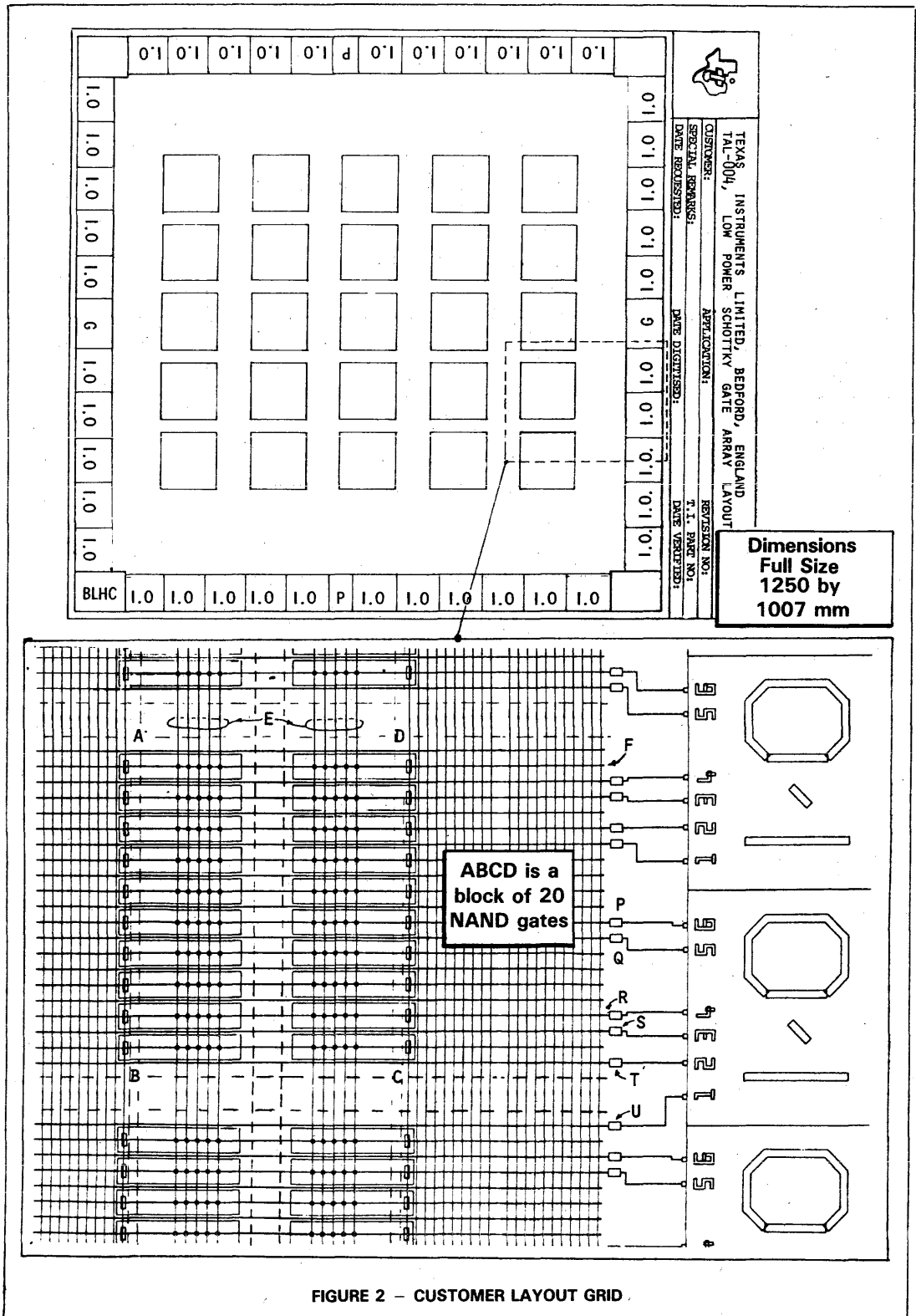
A reproduction of the translucent user layout grid is shown in Figure 2. The lower level of metal interconnect on the finished device is called Metal 1 and is specified on the grid by marking the vertical lines and the upper level, Metal 2, is similarly designated by marking the horizontal lines. Electrical connection between the two layers is made by so called "vias" and to underlying gates by "contacts." Layout design rules govern the location of interconnections, vias, and contacts and these are clearly explained in the design manual.

Standard logic functions and I/O buffers are specified by attaching cut outs to the user grid. Symbols for the logic function overlays (macros) and I/O's are supplied on a sheet similar to the user grid.

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TAL-002, TAL-004  
 LOW POWER SCHOTTKY  
 LOGIC ARRAYS



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**TAL-002, TAL-004  
LOW POWER SCHOTTKY  
LOGIC ARRAYS**

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ .....	7 V
Input voltage .....	7 V
Operating junction temperature .....	150°C
Operating free-air temperature range .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C

recommended operating conditions

PARAMETER	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$	4.75	5	5.25	V
High-level output current, $I_{OH}$			-400	$\mu$ A
Low-level output current, $I_{OL}$			8	mA
Input rise time, $t_r(IN)$	2.4		30	ns
Operating free-air temperature	0		70	°C

**INTERNAL FOUR-INPUT NAND GATES**

recommended operating conditions

PARAMETER	MIN	TYP	MAX	UNIT
Fan-in			8	Gate Loads
Fan-out			8	Gate Loads

switching characteristics over full range of recommended operating conditions

PARAMETER	MIN	TYP	MAX	UNIT	
$t_p$ Propagation delay time	1-input NAND gate	3.5	4.5	5	ns
	2-input NAND gate	4	5.3	5.8	
	3-input NAND gate	4.5	6.1	6.6	
	4-input NAND gate	5	6.9	7.4	
$\Delta t_p$ increased propagation delay time per inch of interconnection track on the user layout grid		0.075		ns	

TABLE 1 - INTERNAL GATE MACROS<sup>†</sup>

SYMBOL	DESCRIPTION		SYMBOL	DESCRIPTION		
DFF	D Flip-Flops	Standard	DEC8	Decoder	3- to 8-line demultiplexer/decoder	
DFFA			Three layouts	MUX4	Multiplexers	4- to 1-line multiplexer
WDF				MUX8		8- to 1-line multiplexer
DFFC		With CLEAR		ADD1	Arithmetic Functions	1-bit full adder
DPC1		With PRESET and CLEAR	EXOR	Exclusive-OR gate		
WDPC			Two layouts	EXNOR		Exclusive-NOR gate
JKPC	J-K Flip-Flop	With PRESET and CLEAR	PAR8	8-bit parity generator		
EFF	Enable Flip-Flops	Standard	MAG4	4-bit magnitude comparator		
DEFF		Driver 2-bit latch	SYCT	Counters	4-bit synchronous up counter	
SEFF		Stacking 2-bit latch	UDCT		4-bit synchronous up/down counter	
DEC4	Decoder	2- to 4-line demultiplexer/decoder	BCD		4-bit synchronous decimal up counter	

<sup>†</sup> For detailed descriptions of the internal gate macros, please refer to the "Low Power Schottky Logic Array Design Manual."

TABLE 2 – INPUT/OUTPUT BUFFERS

SYMBOL	DESCRIPTION	SYMBOL	DESCRIPTION
BF1A	True and Complement Inputs	BF5A	Two-Input Open-Collector Inverting Outputs
BF1B		BF5B	
BF2A	Two-Input NAND Output	BF6A	Inverting Schmitt Trigger Input
BF3A	Standard Input/Output	BF7A	Power-Up Reset Circuit
BF4A	Inverting Three-State Output		

electrical characteristics over full ranges of recommended operating conditions

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
I <sub>CC</sub>	Supply current	BF1A	V <sub>CC</sub> = 4.75 V to 5.25 V, T <sub>A</sub> = 0°C to 70°C, See note 1	1.38	1.95		mA	
		BF2A, BF5A		1.38	1.95			
		BF3A		Output enabled	3.18	4.5		
				Output disabled	3.3	4.8		
		BF4A		Output enabled	2.93	4.15		
				Output disabled	3.15	4.45		
		BF6A		1.46	2.1			
		BF7A		0.68	1			

NOTE 1: Values given assume that the buffer spends equal time in each of its logic states; please refer to the "Low Power Schottky Logic Array Design Manual" for more information.

electrical characteristics over recommended operating temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High-level output voltage		I <sub>OH</sub> = -400 μA, V <sub>CC</sub> = 4.75 V	2.7	3.35		V
V <sub>OL</sub>	Low-level output voltage		I <sub>OL</sub> = 8 mA, V <sub>CC</sub> = 4.75 V		0.35	0.5	V
I <sub>OL</sub>	Low-level output current, open-collector, internal pullup	BF5B	V <sub>O</sub> = 0.4 V	0.35		0.8	mA
I <sub>OS</sub>	Short-circuit output current		V <sub>CC</sub> = 5.25 V	-20	-60	-100	mA
I <sub>OZH</sub>	Off-state output current, high-level voltage applied	BF4A	V <sub>CC</sub> = 5.25 V, V <sub>O</sub> = 2.7 V			20	μA
I <sub>OZL</sub>	Off-state output current, low-level voltage applied	BF4A	V <sub>CC</sub> = 5.25 V, V <sub>O</sub> = 0.4 V			-20	μA

**TAL-002, TAL-004  
LOW POWER SCHOTTKY  
LOGIC ARRAYS**

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
V <sub>IH</sub>	High-level input voltage		2			V	
V <sub>IL</sub>	Low-level input voltage				0.8	V	
V <sub>IK</sub>	Input clamp voltage	I <sub>K</sub> = -18 mA			-1.5	V	
V <sub>T+</sub>	Positive-going threshold voltage	BF6A	1.4	1.68	1.9	V	
V <sub>T-</sub>	Negative-going threshold voltage	BF6A	0.5	0.84	1	V	
I <sub>IH</sub>	High-level input current	BF1A, BF6A	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 2.7 V			20	μA
		BF3A				40	
		BF1B	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 2.7 V			-400	
I <sub>IL</sub>	Low-level input current	BF1A, BF3A, BF6A	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 0.4 V			-200	μA
		BF1B				-850	
I <sub>I</sub>	Input current	BF1B	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 5.5 V			100	μA
		All others	V <sub>CC</sub> = 5.25 V, V <sub>I</sub> = 7 V			100	

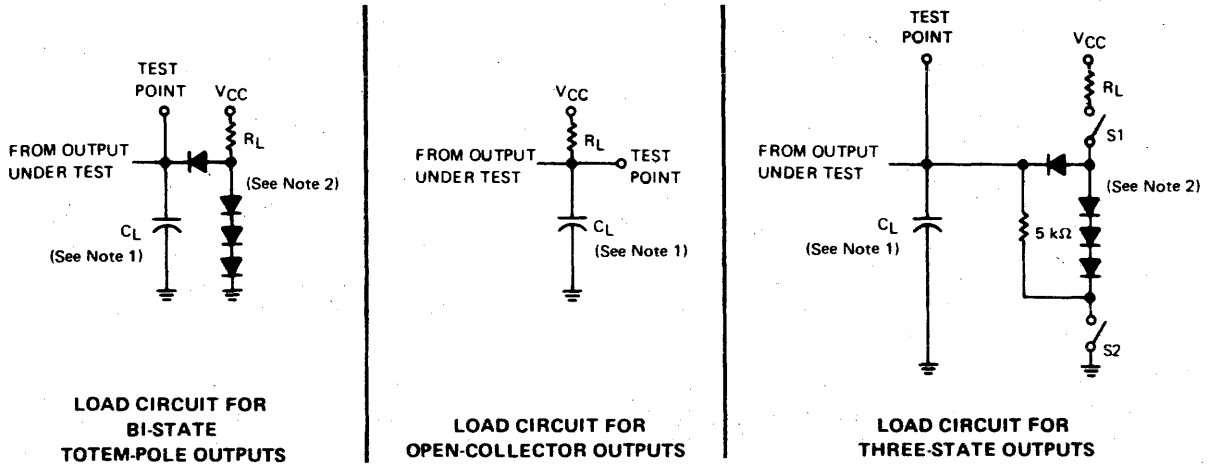
switching characteristics over full ranges of recommended operating conditions

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT	
t <sub>PLH</sub>	Propagation delay, low-to-high level	BF1A True	V <sub>CC</sub> = 4.75 V to 5.25 V, T <sub>A</sub> = 0°C to 70°C, See Figure 3	5	10	15	ns	
		BF1A Complement		4	7	10		
		BF2A Output port		9	11	13		
		BF3A Input mode		True	5	10		15
				Complement	4	7		10
		BF3A Output port		8	11	14		
		BF4A Output port		8	11	14		
		BF5A Output port		10	15	25		
BF6A Complement array input	9	12	15					
t <sub>PHL</sub>	Propagation delay, high-to-low level	BF1A True	V <sub>CC</sub> = 4.75 V to 5.25 V, T <sub>A</sub> = 0°C to 70°C, See Figure 3	5	9	12	ns	
		BF1A Complement		3	5	7		
		BF2A Output port		5	7	9		
		BF3A Input mode		True	5	9		12
				Complement	3	5		7
		BF3A Output port		7	10	13		
		BF4A Output port		7	10	13		
		BF5A Output port		5	7	9		
BF6A Complement array input	8	10	12					
t <sub>PZH</sub>	Output enable time to high level	BF3A Output port	V <sub>CC</sub> = 4.75 V to 5.25 V, T <sub>A</sub> = 0°C to 70°C, See Figure 3	10	14	18	ns	
		BF4A Output port		10	14	18		
t <sub>PZL</sub>	Output enable time to low level	BF3A Output port		10	14	18	ns	
		BF4A Output port		10	14	18		
t <sub>PHZ</sub>	Output disable time to high level	BF3A Output port		12	17	22	ns	
		BF4A Output port		12	17	22		
t <sub>PLZ</sub>	Output disable time to low level	BF3A Output port		12	17	22	ns	
		BF4A Output port		12	17	22		

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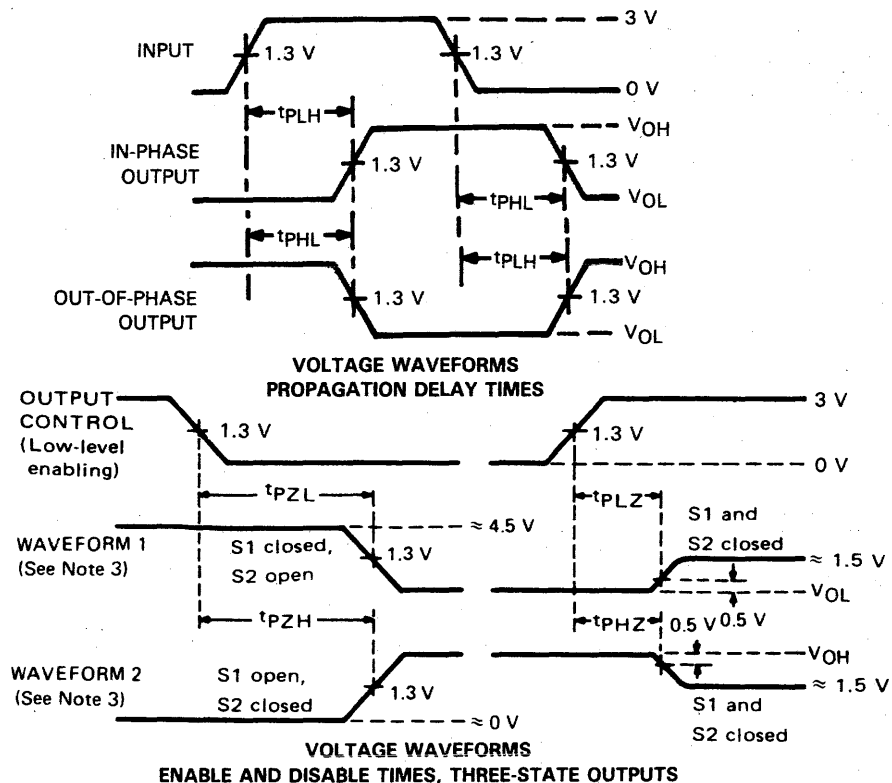
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PARAMETER MEASUREMENT INFORMATION†



† I/O cells only.  
 NOTES: 1.  $C_L$  includes probe and jig capacitance.  
 2. All diodes are 1N916 or 1N3064.

FIGURE 3 - LOAD CIRCUITS



NOTES: 3. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.  
 4. In the examples above, the phase relationships between inputs and outputs have been chosen arbitrarily.  
 5. All input pulses are supplied by generators having the following characteristics:  $t_r \leq 15$  ns,  $t_f \leq 15$  ns, PRR  $\leq 1$  MHz,  $Z_{out} = 50 \Omega$ .  
 6. When measuring propagation delay times of 3-state outputs, switches S1 and S2 are closed.

FIGURE 4 - SWITCHING TIMES

Texas Instruments reserves the right to make changes at any time in order to improve design and to supply the best product possible.



- **High-Performance Master Logic Arrays**
  - Designed for Automated Layout
  - Mask Programmable
  - Supported by TI Logic Array Design System (TILADS\*)
- **Two Master Arrays**
- **High-Performance Schottky Transistor Logic (STL) Design**
  - 2.5 ns Typical Gate Propagation Delay at 600  $\mu$ W
  - 80-MHz Toggle Rate
- **Low-Power Schottky TTL Compatible Input and Output Buffers**
  - Non-Inverting Inputs
  - Choice of 3-State, Totem-Pole and Open-Collector Outputs

	Internal Gates	Typical Usable Internal Gates	I/O Buffers
TAT008	1008	800	104
TAT004	540	430	76

**description**

The TAT004 and TAT008 are a pair of high-performance logic arrays, using Schottky Transistor Logic (STL) technology. The arrays can be mask configured to fulfill unique logic requirements, allowing for efficient implementation of custom IC functions, SSI or MSI logic replacement, and, in many cases, complete board replacement.

The master array employs a cellular organization with channels dedicated to intercellular connections. The internal cells contain 18 high-performance STL gates, and the peripheral cells contain low-power Schottky TTL (TTL-LS) compatible input and output buffers.

All Texas Instruments TAT Series logic arrays are supported by a totally integrated software design utility, which allows the user to encode his logic design and test pattern set for simulation, analysis, and verification of the design prior to release to TI for layout and prototyping. The computer-automated layout system has a recommended 80% gate utilization level.

**package options**

The TAT004 and the TAT008 are available in the packages indicated by check marks in the table below.

TAT004	40-PIN	64-PIN		
Dual-in-line plastic (N) packages	✓	✓		
TAT008	40-PIN	68-PIN	84-PIN	108-PIN
Dual-in-line plastic (N) packages	✓	-	-	-
Ceramic chip carriers	-	-	✓	-
Plastic chip carriers	-	✓	-	-
Pin grid arrays	-	-	✓	✓

**input/output buffer options**

The TAT004 and TAT008 are provided with a library of input/output circuits for maximum design flexibility. The buffers are arranged around the periphery of the array in cells, called buffer cells. Each buffer cell contains two bond pad locations plus the basic components needed to mask program the cell into various TTL low-power Schottky-compatible buffers. A buffer cell can contain two input buffers, but if a cell contains one of the available output buffer types, then only one input buffer can share the buffer cell with it. Table 1 lists the available buffers along with their label as used by the TI Logic Array Design System.

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# TAT004, TAT008 STL LOGIC ARRAYS

TABLE 1 - TAT004 AND TAT008 BUFFERS

SYMBOL	DESCRIPTION	SYMBOL	DESCRIPTION
SNI	Noninverting Input	SOT	inverting 3-State Output with Active Low Enable
SNI2	Dual Noninverting Inputs		
SAND	AND Input	SBD	Bidirectional Buffer with Inverting 3-State Output, Active Low Enable
SAND2	Dual AND Inputs		
SO	Inverting Totem-Pole Output	SNIO	Bidirectional Buffer with Inverting Open-Collector Output
SOC	Inverting Open-Collector Output		

**absolute maximum ratings over operating free-air temperature range**

Supply voltage: $V_{CC1}$ (see Note 1)	7 V
$V_{CC2}$	3 V
Input voltage	7 V
Output voltage	7 V
Operating free-air temperature range	0°C to 70°C
Operating junction temperature	150°C
Storage temperature range	-65°C to 150°C

NOTE 1: All voltages are with respect to network ground terminal.

**recommended operating conditions**

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC1}$	4.5	5	5.5	V
Supply voltage, $V_{CC2}$	1.8	2	2.2	V
Fan-in (internal gate)			8	
Fan-out (internal gate)			4	
High-level output current, $I_{OH}$			-400	$\mu$ A
Low-level output current, $I_{OL}$			8	mA
Transition time at any input, $t_T$	2.4		30	ns
Operating free-air temperature range, $T_A$	0		70	°C

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electrical characteristics over recommended ranges of supply voltages and operating free-air temperature

PARAMETER	TEST CONDITIONS	MIN	TYP†	MAX	UNIT
V <sub>IH</sub> High-level input voltage		2			V
V <sub>IL</sub> Low-level input voltage				0.8	V
V <sub>IK</sub> Input clamp voltage	I <sub>I</sub> = -18 mA			-1.5	V
V <sub>OH</sub> High-level output voltage (see Note 2)	I <sub>OH</sub> = -400 μA	2.5	3.4		V
V <sub>OL</sub> Low-level output voltage	I <sub>OL</sub> = 8 mA			0.5	V
I <sub>OZH</sub> Off-state output current, high-level voltage applied (see Note 3)	V <sub>O</sub> = 2.4 V			20	μA
I <sub>OZL</sub> Off-state output current, low-level voltage applied (see Note 3)	V <sub>O</sub> = 0.4 V			-20	μA
I <sub>I</sub> Input current at maximum input voltage	V <sub>I</sub> = 7 V			0.1	mA
I <sub>IH</sub> High-level input current	V <sub>I</sub> = 2.7 V			20	μA
I <sub>IL</sub> Low-level input current	V <sub>I</sub> = 0.4 V			0.8	mA
I <sub>OS</sub> Short-circuit output current (see Note 2 and 4)	V <sub>CC1</sub> = 5.5 V	-20		-100	mA
I <sub>OH</sub> High-level output current (see Note 5)	V <sub>OH</sub> = 5.5 V			250	μA

† All typical values are at T<sub>A</sub> = 25°C, V<sub>CC1</sub> = 5 V, V<sub>CC2</sub> = 2 V.

NOTES: 2. Does not apply to SOC or SNIO buffers.

3. Applies only to SOT buffer.

4. Not more than one output should be shorted at a time, and duration of the short circuit should not exceed one second.

5. Applies only to SOC and SNIO buffers.

switching characteristics of internal gate over full range of operating conditions

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>pd</sub> Propagation delay time per gate, fan-in = 1		1.5	2.5	3.5	ns
Δt <sub>PLH</sub> Increased propagation delay time for each additional fan-in			0.27	0.40	ns
Δt <sub>PHL</sub> Increased propagation delay time for each additional fan-in			0.13	0.20	ns
Δt <sub>PLH(M1)</sub> † Additional propagation delay time for each mil of metalized interconnect pattern			0.017	0.040	ns
Δt <sub>PLH(M2)</sub> † Additional propagation delay time for each mil of metalized interconnect pattern			0.015	0.035	ns
Δt <sub>PHL(M)</sub> † Additional propagation delay time for each mil of metalized interconnect pattern				0.003	ns

† M1 = first-level metal, M2 = second-level metal, and M = first- or second-level metal.

switching characteristics of input buffer over full range of operating conditions

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	See Note 6	1	2	4	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-level output		2	3.5	6	ns

switching characteristics of tristate and totem-pole output buffers over full range of operating conditions

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> Propagation delay time, low-to-high-level output	C <sub>L</sub> = 50 pF, R <sub>L</sub> = 2 kΩ, See Note 6	4	7	11	ns
t <sub>PHL</sub> Propagation delay time, high-to-low-level output		6	12	19	ns
t <sub>PHZ</sub> Disable time from high level	C <sub>L</sub> = 15 pF, R <sub>L</sub> = 2 kΩ, See Note 6	8	14	20	ns
t <sub>PLZ</sub> Disable time from low level		6	10	15	ns
t <sub>pZH</sub> Enable time to high level		4	8	12	ns
t <sub>pZL</sub> Enable time to low level		5	10	16	ns

NOTE 6: Refer to the TAL-002, TAL-004 Data Sheet for load circuits and switching times.

# TAT004, TAT008 STL LOGIC ARRAYS

switching characteristics of open-collector output buffers over full range of operating conditions

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay, low-to-high-level output	C <sub>L</sub> = 50 pF, R <sub>L</sub> = 2 kΩ,	11	20	31	ns
t <sub>PHL</sub>	Propagation delay, high-to-low-level output	See Note 6	9	21	33	ns

NOTE 6: Refer to the TAL-002, TAL-004 Data Sheet for load circuits and switching times.

power dissipation versus package, T<sub>A</sub> = 70°C (still air, no heat sinking, T<sub>J</sub> = 150°C maximum)

PACKAGE	θ <sub>J-A</sub>	POWER DISSIPATION
	(°C/WATT)	MAX (WATTS)
40-pin plastic dual-in-line	65	1.23
64-pin plastic dual-in-line	45	1.78
68-pin plastic chip carrier	46	1.74
84-pin ceramic chip carrier	39	2.05
84-pin ceramic pin grid array	39	2.05
108-pin ceramic pin grid array	38	2.11

power dissipation versus buffer type (see Note 7)

BUFFER	POWER DISSIPATION (mW)				
	T <sub>J</sub> = 25°C		T <sub>J</sub> = 125°C		
	TYP <sup>†</sup>	MAX <sup>‡</sup>	TYP <sup>†</sup>	MAX <sup>‡</sup>	
SNI	3.60	5.94	3.00	4.94	
SAND	3.60	5.94	3.00	4.94	
SO	8.08	13.31	6.67	10.95	
SOC	5.83	9.63	4.81	7.91	
SBD	Output enabled	14.59	23.98	12.15	19.75
	Output disabled	16.90	27.77	13.94	22.84
SOT	Output enabled	11.00	18.06	9.06	14.82
	output disabled	11.82	19.34	9.66	15.77
SNIO	Output active	9.43	15.57	7.81	12.85
Internal gate (V <sub>CC2</sub> )		0.668	1.12	0.574	0.959

NOTE 7: All power dissipation values are with 50% duty cycle.

<sup>†</sup> Typical values are with nominal resistors, V<sub>CC1</sub> = 5 V, and V<sub>CC2</sub> = 2 V.

<sup>‡</sup> Maximum values are with minimum resistors, V<sub>CC1</sub> = 5.5 V, and V<sub>CC2</sub> = 2.2 V.

## PARAMETER MEASUREMENT INFORMATION

Refer to the TAL-002, TAL-004 Data Sheet for load circuits and switching times.

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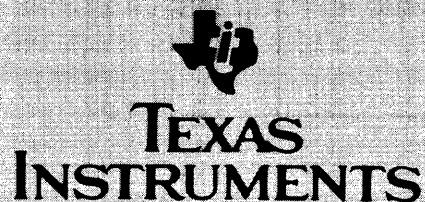


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# Field Programmable Logic

Product Guide



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**NUMERICAL INDEX**  
**FIELD PROGRAMMABLE LOGIC ARRAYS**

DEVICE	FUNCTION	AVAILABILITY
PL16L8	FPL Fixed-OR Arrays	NOW
PL16R4		NOW
PL16R6		NOW
PL16R8		NOW
PL20L8	FPL Fixed-OR Arrays	▲
PL20R4		▲
PL20R6		▲
PL20R8		▲
PL839	14 X 32 X 6 Field Programmable Logic Arrays	▲
PL840	14 X 32 X 6 Field Programmable Logic Arrays	▲
PLR19L8	FPL Registered-Input Fixed-OR Arrays	▲
PLR19R4		▲
PLR19R6		▲
PLR19R8		▲
PLT19L8	FPL Latched-Input Fixed-OR Arrays	▲
PLT19R4		▲
PLT19R6		▲
PLT19R8		▲
<p>▲ Planned for future announcement. Contact TI field sales office or authorized TI distributor for status.</p>		

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**'PL16L8**

functional block diagram

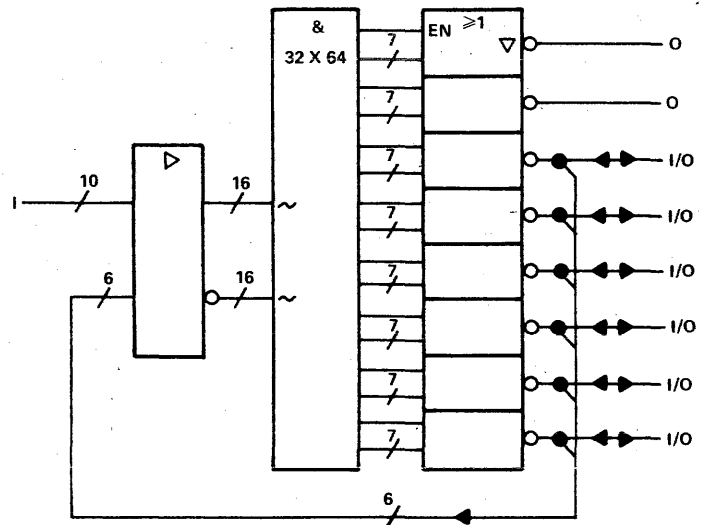
**FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS**

- Octal 16-input AND-OR-INVERT gate array

pin assignments

J, N PACKAGES			
1	I	11	I
2	I	12	O
3	I	13	I/O
4	I	14	I/O
5	I	15	I/O
6	I	16	I/O
7	I	17	I/O
8	I	18	I/O
9	I	19	O
10	GND	20	V <sub>CC</sub>

SN54PL16L8 (J)      SN74PL16L8 (J,N)



~denotes fused inputs

**'PL16R4**

functional block diagram

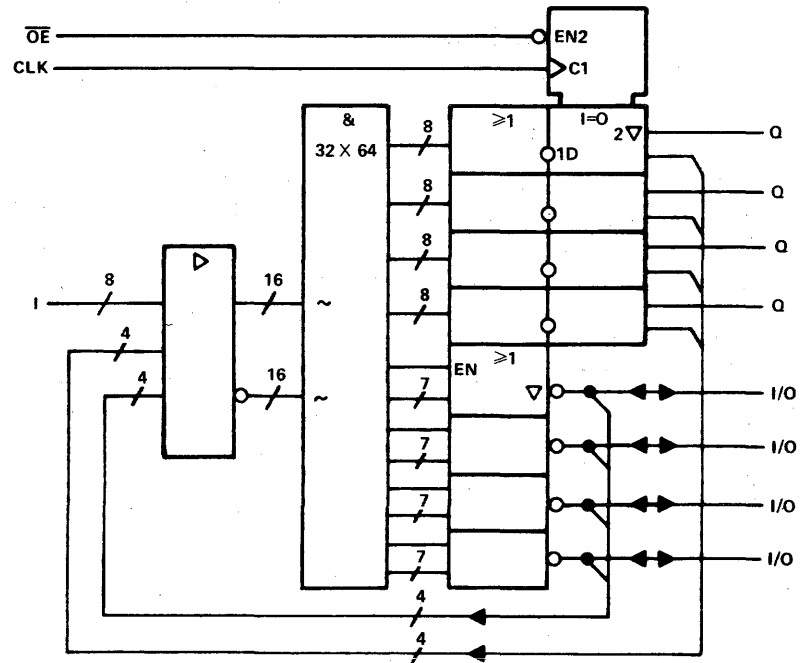
**FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS**

- Quad 16-input registered AND-OR gate array

pin assignments

J, N PACKAGES			
1	CLK	11	$\overline{OE}$
2	I	12	I/O
3	I	13	I/O
4	I	14	Q
5	I	15	Q
6	I	16	Q
7	I	17	Q
8	I	18	I/O
9	I	19	I/O
10	GND	20	V <sub>CC</sub>

SN54PL16R4 (J)      SN74PL16R4 (J,N)



~denotes fused inputs

Refer to data sheet in this section for additional information.

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

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'PL16R6

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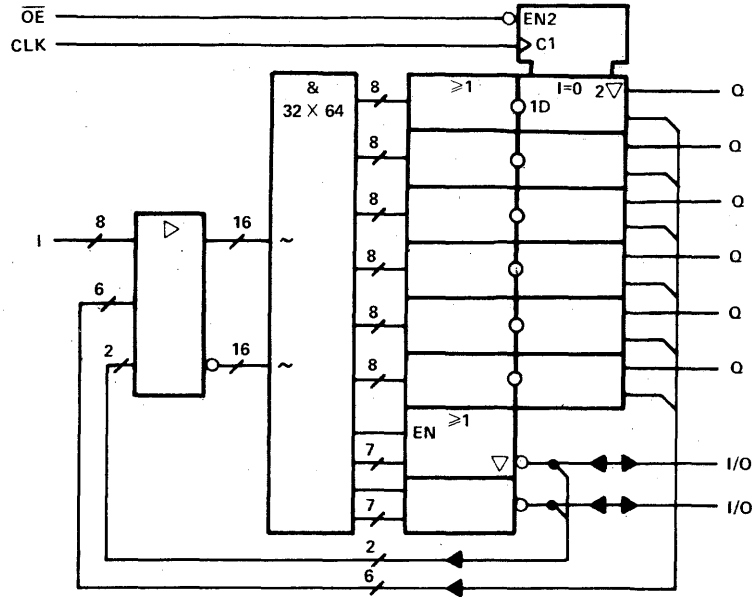
- Hex 16-input registered AND-OR gate array

pin assignments

J, N PACKAGES		
1	CLK	11 $\overline{OE}$
2	I	12 I/O
3	I	13 Q
4	I	14 Q
5	I	15 Q
6	I	16 Q
7	I	17 Q
8	I	18 Q
9	I	19 I/O
10	GND	20 $V_{CC}$

SN54PL16R6 (J) SN74PL16R6 (J,N)

functional block diagram



~denotes fused inputs

'PL16R8

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

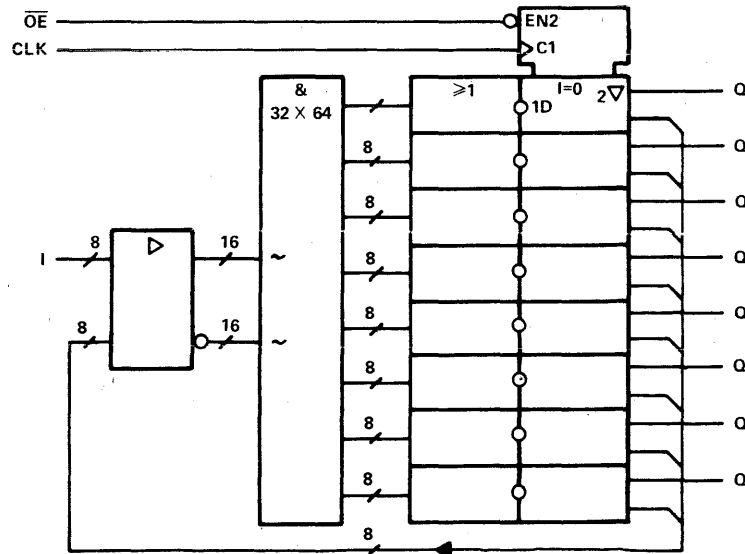
- Octal 16-input registered AND-OR gate array

pin assignments

J, N PACKAGES		
1	CLK	11 $\overline{OE}$
2	I	12 Q
3	I	13 Q
4	I	14 Q
5	I	15 Q
6	I	16 Q
7	I	17 Q
8	I	18 Q
9	I	19 Q
10	GND	20 $V_{CC}$

SN54PL16R8 (J) SN74PL16R8 (J,N)

functional block diagram



~denotes fused inputs

Refer to data sheet in this section for additional information.

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

### 'PL20L8

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

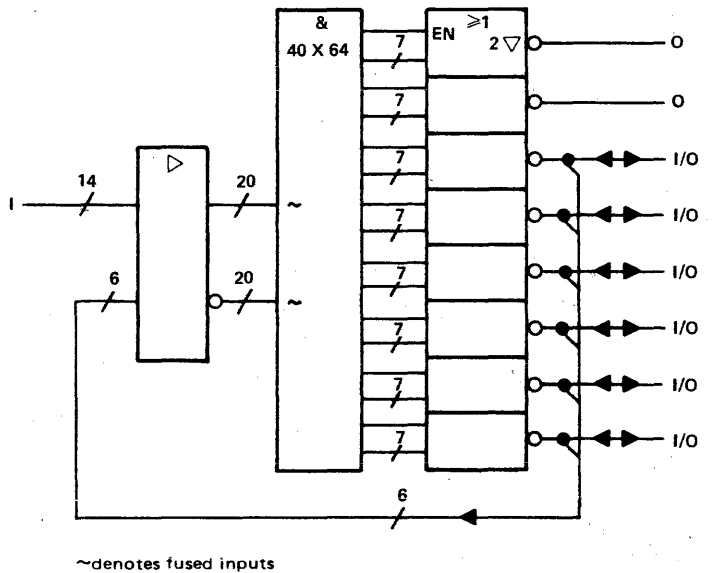
• Octal 20-input AND-OR-INVERT gate array

pin assignments

JT, NT PACKAGES			
1	I	13	I
2	I	14	I/PRELOAD
3	I	15	O
4	I	16	I/O
5	I	17	I/O
6	I	18	I/O
7	I	19	I/O
8	I	20	I/O
9	I	21	I/O
10	I	22	O
11	I	23	I
12	GND	24	VCC

SN54PL20L8 (JT) SN74PL20LB (JT,NT)

functional block diagram



### 'PL20R4

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

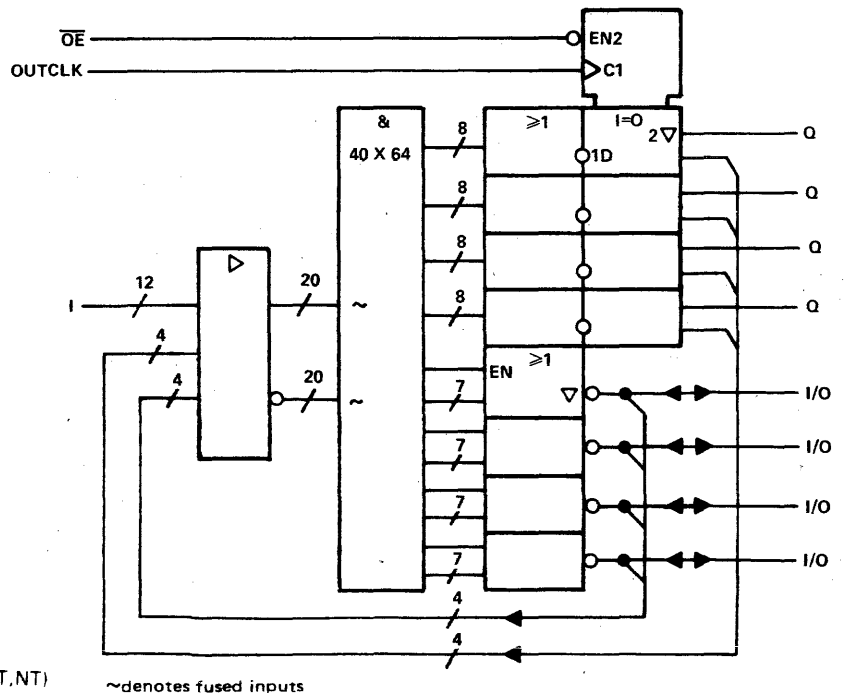
• Quad 20-input registered AND-OR gate array

pin assignments

JT, NT PACKAGES			
1	OUTCLK	13	OE
2	I	14	I/PRELOAD
3	I	15	I/O
4	I	16	I/O
5	I	17	Q
6	I	18	Q
7	I	19	Q
8	I	20	Q
9	I	21	I/O
10	I	22	I/O
11	I	23	I
12	GND	24	VCC

SN54PL20R4 (JT) SN74PL20R4 (JT,NT)

functional block diagram



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'PL20R6

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

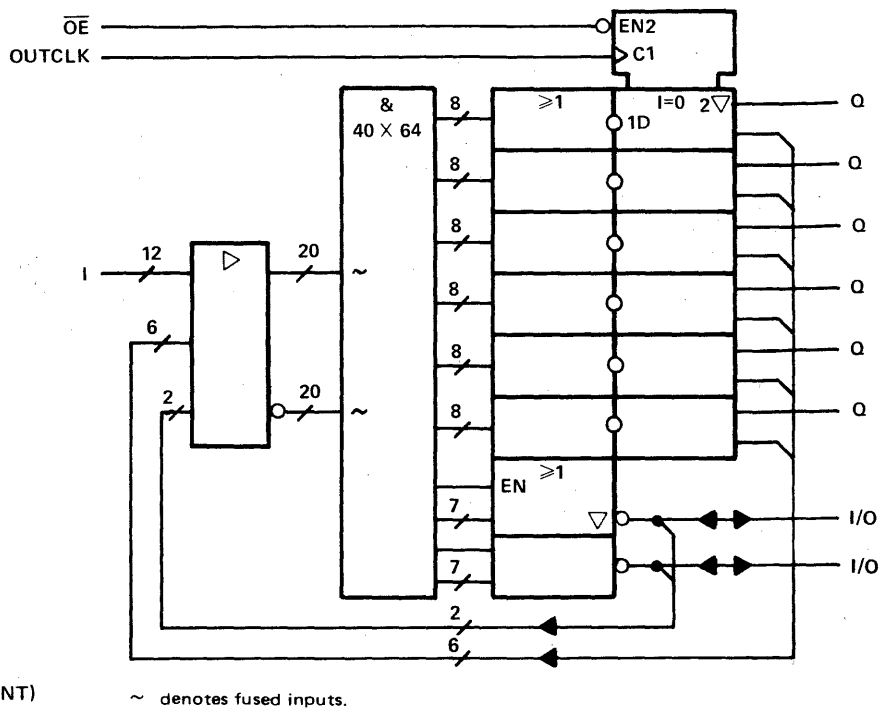
- Hex 20-input registered AND-OR gate array.

pin assignments

JT, NT PACKAGES		
1	OUT CLK	24 V <sub>CC</sub>
2	I	23 I
3	I	22 I/O
4	I	21 Q
5	I	20 Q
6	I	19 Q
7	I	18 Q
8	I	17 Q
9	I	16 Q
10	I	15 I/O
11	I	14 I/PRELOAD
12	GND	13 $\overline{OE}$

SN54PL20R6 (JT) SN74PL20R6 (JT, NT)

functional block diagram



'PL20R8

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

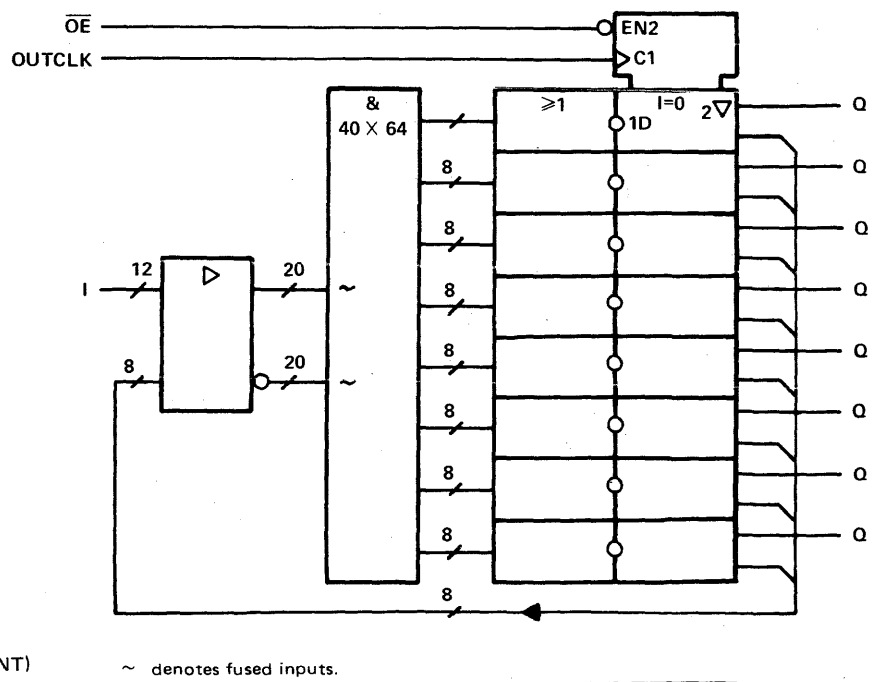
- Octal 20-input registered AND-OR gate array.

pin assignments

JT, NT PACKAGES		
1	OUT CLK	24 V <sub>CC</sub>
2	I	23 I
3	I	22 Q
4	I	21 Q
5	I	20 Q
6	I	19 Q
7	I	18 Q
8	I	17 Q
9	I	16 Q
10	I	15 Q
11	I	14 I/PRELOAD
12	GND	13 $\overline{OE}$

SN54PL20R8 (JT) SN74PL20R8 (JT, NT)

functional block diagram



Refer to data sheet in this section for additional information.

# 'PL839, PL840

## FIELD-PROGRAMMABLE LOGIC ARRAYS

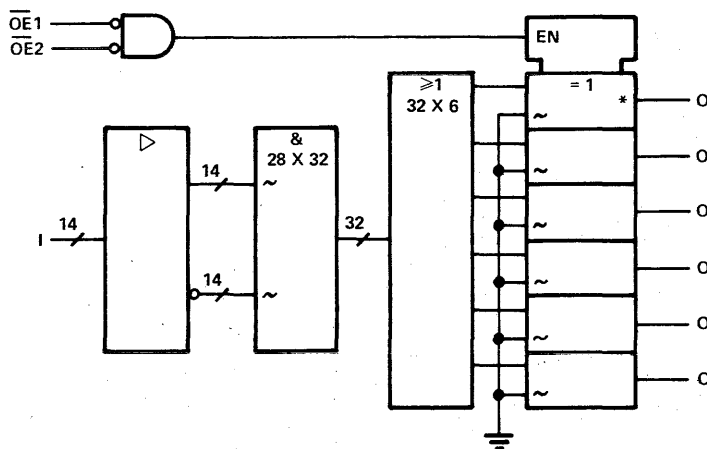
- 'PL839 — three-state outputs
- 'PL840 — open-collector outputs
- Programmable output polarity

### pin assignments

J, JT, NT PACKAGES			
1	$\overline{OE1}$	13	$\overline{OE2}$
2	I	14	O
3	I	15	O
4	I	16	O
5	I	17	I
6	I	18	I
7	I	19	I
8	I	20	I
9	O	21	I
10	O	22	I
11	O	23	I
12	GND	24	V <sub>CC</sub>

SN54PL839 (J)      SN74PL839 (JT,NT)  
 SN54PL840 (J)      SN74PL840 (JT,NT)

functional block diagram



~ denotes fused inputs.  
 \* 'PL839 has 3-state ( $\nabla$ ) outputs; 'PL840 has open-collector ( $\diamond$ ) outputs.

nc — no internal connection.

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'PLR19L8

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

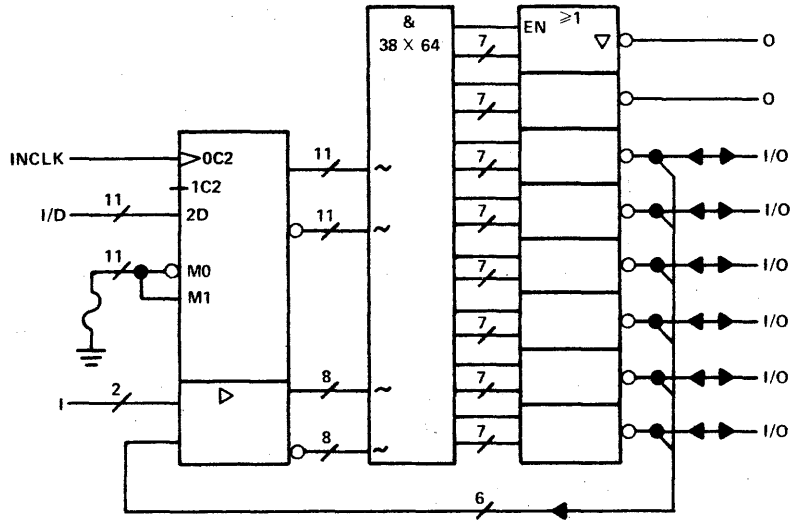
- Octal 19-input registered AND-OR-INVERT gate array

pin assignments

JT, NT PACKAGES	
1 I	13 I
2 I/D	14 INCLK/PRELOAD
3 I/D	15 O
4 I/D	16 I/O
5 I/D	17 I/O
6 I/D	18 I/O
7 I/D	19 I/O
8 I/D	20 I/O
9 I/D	21 I/O
10 I/D	22 O
11 I/D	23 I/D
12 GND	24 V <sub>CC</sub>

SN54PLR19L8 (JT) SN74PLR19L8 (JT,NT)

functional block diagram



'PLR19R4

FIELD-PROGRAMMABLE LOGIC, FIXED-OR ARRAYS

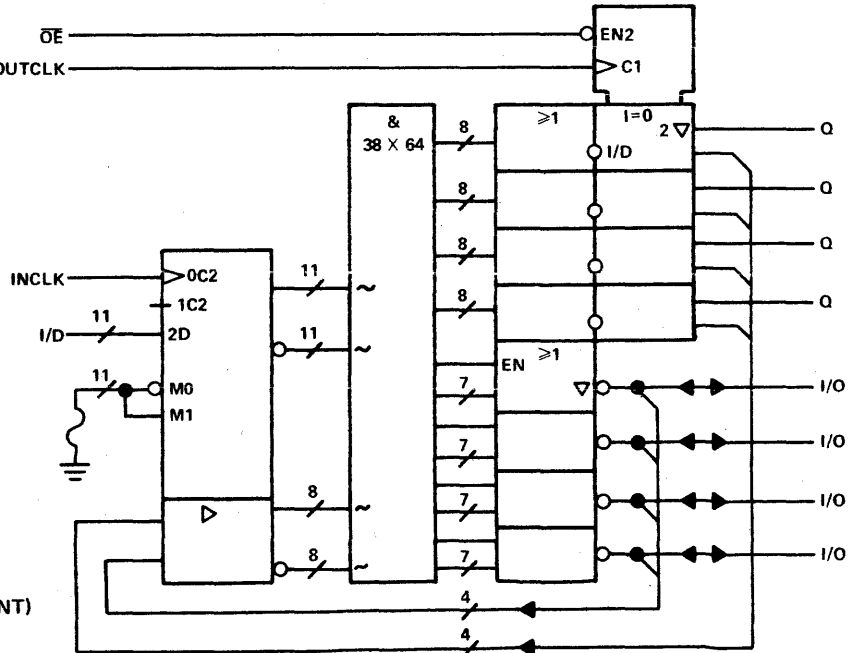
- Quad 19-input registered AND-OR gate array

pin assignments

JT, NT PACKAGES	
1 OUTCLK	13 $\overline{OE}$
2 I/D	14 INCLK/PRELOAD
3 I/D	15 I/O
4 I/D	16 I/O
5 I/D	17 Q
6 I/D	18 Q
7 I/D	19 Q
8 I/D	20 Q
9 I/D	21 I/O
10 I/D	22 I/O
11 I/D	23 I/D
12 GND	24 V <sub>CC</sub>

SN54PLR19R4 (JT) SN74PLR19R4 (JT,NT)

functional block diagram



Refer to data sheet in this section for additional information.

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

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**'PLR19R6**

FIELD-PROGRAMMABLE  
LOGIC, FIXED-OR ARRAYS

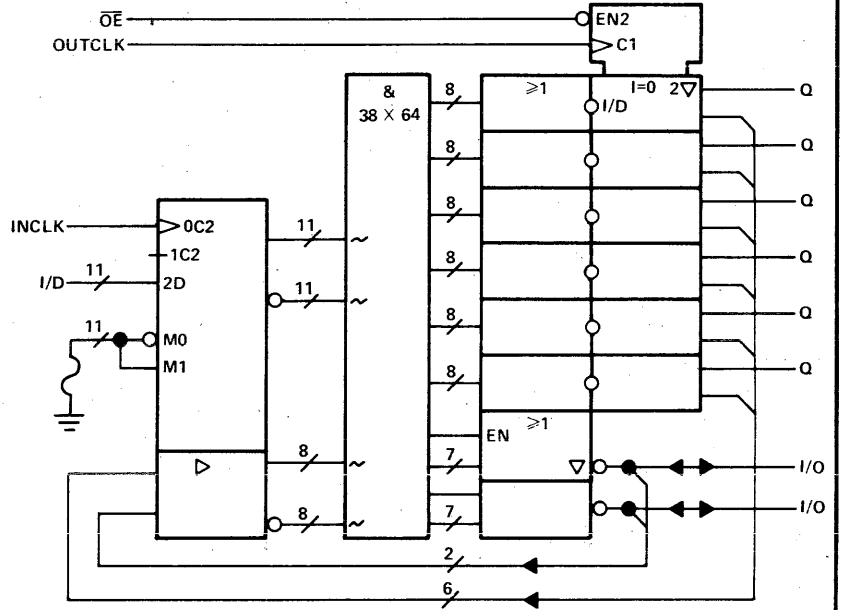
- Hex 19-input registered AND-OR gate array

pin assignments

JT, NT PACKAGES	
1 OUTCLK	13 $\overline{OE}$
2 I/D	14 INCLK/PRELOAD
3 I/D	15 I/O
4 I/D	16 Q
5 I/D	17 Q
6 I/D	18 Q
7 I/D	19 Q
8 I/D	20 Q
9 I/D	21 Q
10 I/D	22 I/O
11 I/D	23 I/D
12 GND	24 $V_{CC}$

SN54PLR19R6 (JT) SN74PLR19R6 (JT,NT)

functional block diagram



**'PLR19R8**

FIELD-PROGRAMMABLE  
LOGIC, FIXED-OR ARRAYS

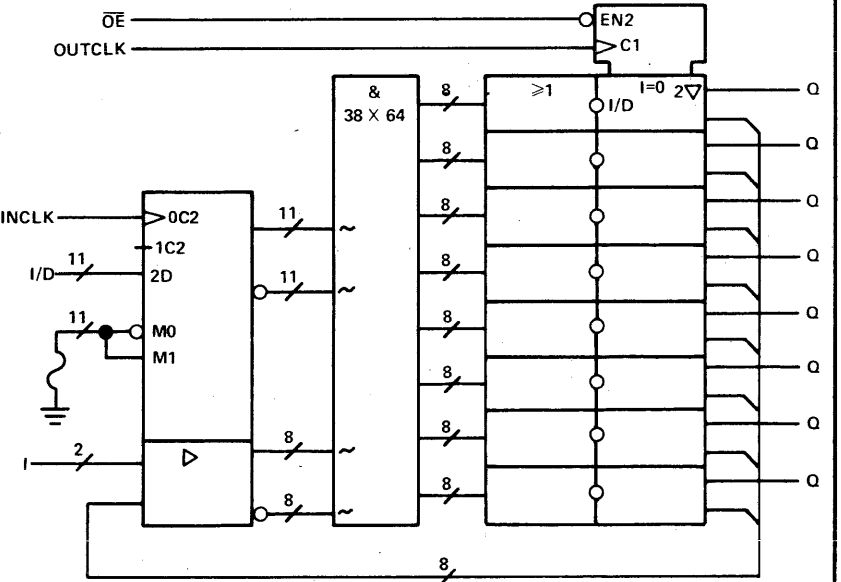
- Octal 19-input registered AND-OR gate array

pin assignments

JT, NT PACKAGES	
1 OUTCLK	13 $\overline{OE}$
2 I/D	14 INCLK/PRELOAD
3 I/D	15 Q
4 I/D	16 Q
5 I/D	17 Q
6 I/D	18 Q
7 I/D	19 Q
8 I/D	20 Q
9 I/D	21 Q
10 I/D	22 Q
11 I/D	23 I/D
12 GND	24 $V_{CC}$

SN54PLR19R8 (JT) SN74PLR19R8 (JT,NT)

functional block diagram



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**'PLT19L8**

FIELD-PROGRAMMABLE  
LOGIC, FIXED-OR ARRAYS

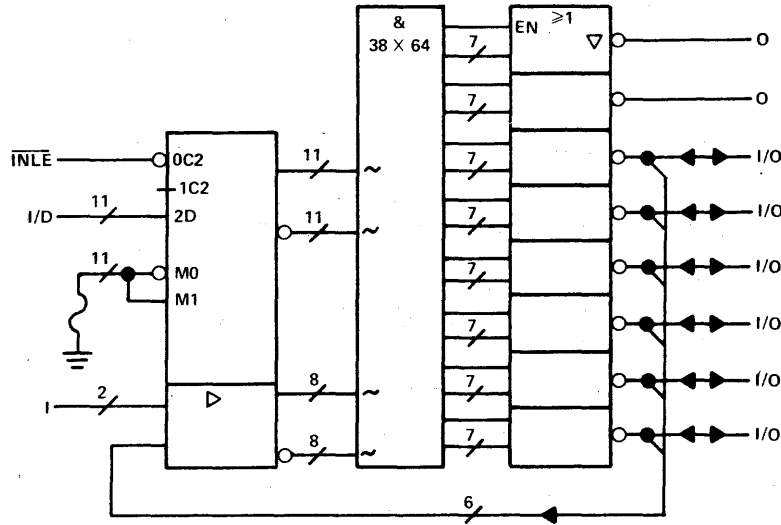
• Octal 19-input latched AND-OR-INVERT gate array

pin assignments

JT, NT PACKAGES	
1 I	13 I
2 I/D	14 INLE/PRELOAD
3 I/D	15 O
4 I/D	16 I/O
5 I/D	17 I/O
6 I/D	18 I/O
7 I/D	19 I/O
8 I/D	20 I/O
9 I/D	21 I/O
10 I/D	22 O
11 I/D	23 I/D
12 GND	24 VCC

SN54PLT19L8 (JT) SN74PLT19L8 (JT,NT)

functional block diagram



**'PLT19R4**

FIELD-PROGRAMMABLE  
LOGIC, FIXED-OR ARRAYS

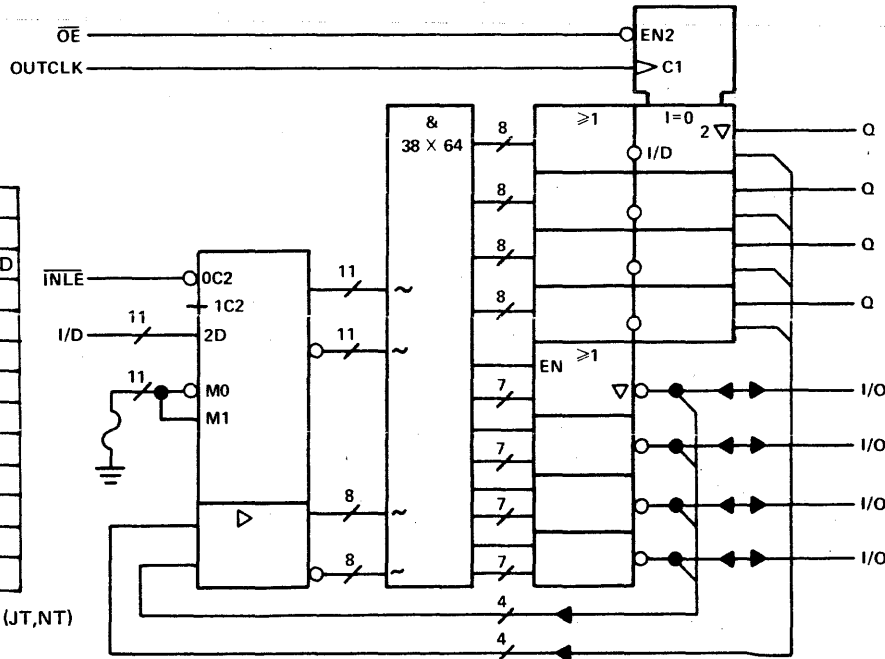
• Quad 19-input registered AND-OR gate array.

pin assignments

JT, NT PACKAGES	
1 OUTCLK	13 OE
2 I/D	14 INLE/PRELOAD
3 I/D	15 I/O
4 I/D	16 I/O
5 I/D	17 Q
6 I/D	18 Q
7 I/D	19 Q
8 I/D	20 Q
9 I/D	21 I/O
10 I/D	22 I/O
11 I/D	23 I/D
12 GND	24 VCC

SN54PLT19R4 (JT) SN74PLT19R4 (JT,NT)

functional block diagram



Refer to data sheet in this section for additional information.

**'PLT19R6**

FIELD-PROGRAMMABLE  
LOGIC, FIXED-OR ARRAYS

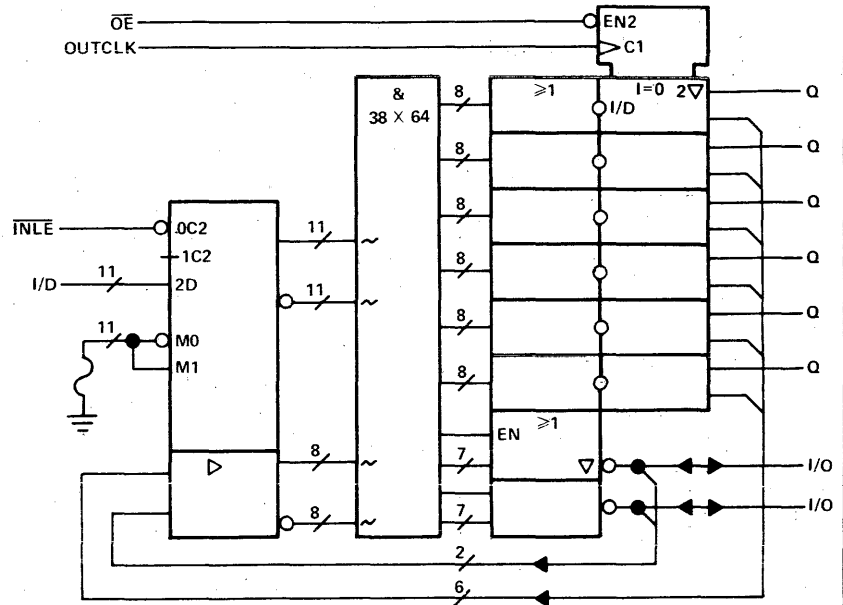
- Hex 19-input latched AND-OR gate array

pin assignments

JT, NT PACKAGES			
1	OUTCLK	13	$\overline{OE}$
2	I/D	14	$\overline{INLE}/\text{PRELOAD}$
3	I/D	15	I/O
4	I/D	16	Q
5	I/D	17	Q
6	I/D	18	Q
7	I/D	19	Q
8	I/D	20	Q
9	I/D	21	Q
10	I/D	22	I/O
11	I/D	23	I/D
12	GND	24	V <sub>CC</sub>

SN54PLT19R6 (JT) SN74PLT19R6 (JT,NT)

functional block diagram



**'PLT19R8**

FIELD-PROGRAMMABLE  
LOGIC, FIXED-OR ARRAYS

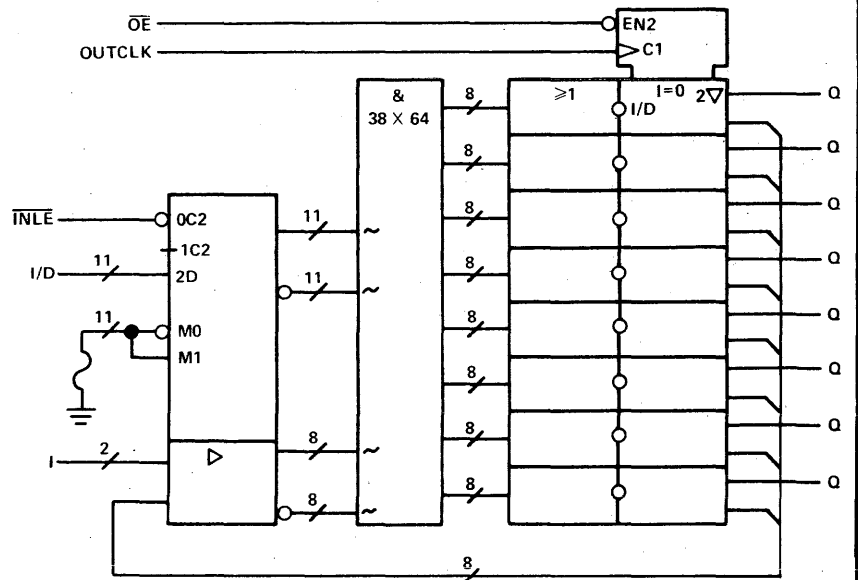
- Octal 19-input latched AND-OR gate array

pin assignments

JT, NT PACKAGES			
1	OUTCLK	13	$\overline{OE}$
2	I/D	14	$\overline{INLE}/\text{PRELOAD}$
3	I/D	15	Q
4	I/D	16	Q
5	I/D	17	Q
6	I/D	18	Q
7	I/D	19	Q
8	I/D	20	Q
9	I/D	21	Q
10	I/D	22	Q
11	I/D	23	I/D
12	GND	24	V <sub>CC</sub>

SN54PLT19R8 (JT) SN74PLT19R8 (JT,NT)

functional block diagram



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# TYPES SN54PL16L8, SN54PL16R4, SN54PL16R6, SN54PL16R8, SN74PL16L8, SN74PL16R4, SN74PL16R6, SN74PL16R8 FIXED-OR ARRAYS

D2705, DECEMBER 1982 - REVISED SEPTEMBER 1983

- Standard 20-Pin, 300-mil Packages
- Choice of operating Speeds
  - 1 Parts . . . 35 MHz Max, Standard power
  - 2 Parts . . . 25 MHz Max, Half power
- Plug-In Compatible with Part Numbers:  
PAL16L8, PAL16R4, PAL16R6, PAL16R8

DEVICE	I INPUTS	3-STATE O OUTPUTS	REGISTERED Q OUTPUTS	I/O PORTS
'PL16L8	10	2	0	6
'PL16R4	8	0	4 (3-state)	4
'PL16R6	8	0	6 (3-state)	2
'PL16R8	8	0	8 (3-state)	0

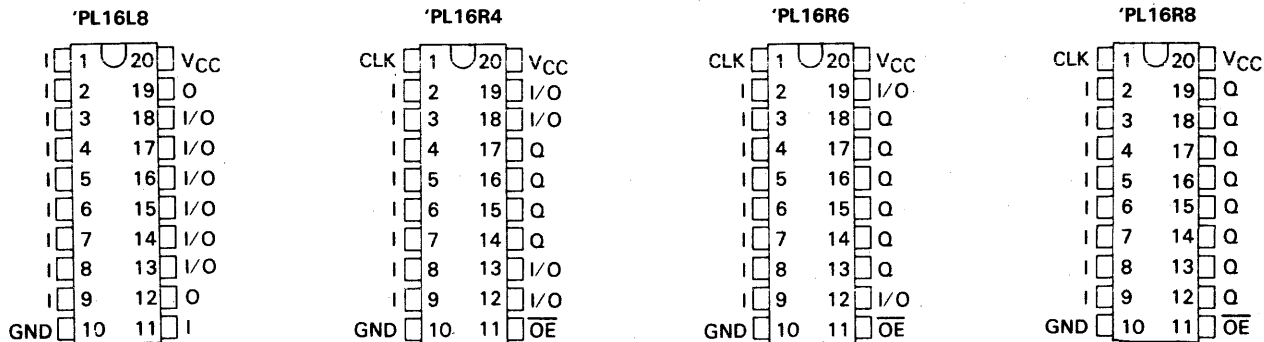
## description

These fixed-OR arrays provide 3-state outputs for bus-oriented systems. They combine Advanced Low-Power Schottky technology with proven titanium-tungsten fuses for reliable, high-performance substitutes for conventional TTL logic. Standard arrays and programmability allow quick design of "custom" functions and more compact boards. The - 1 and - 2 parts offer a choice of operating frequency, switching times, and power dissipation.

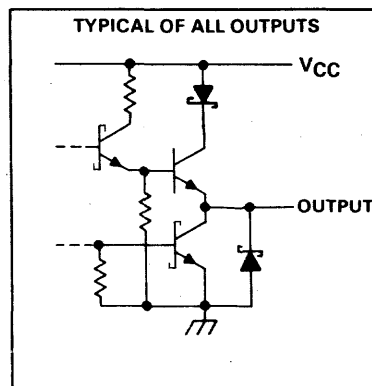
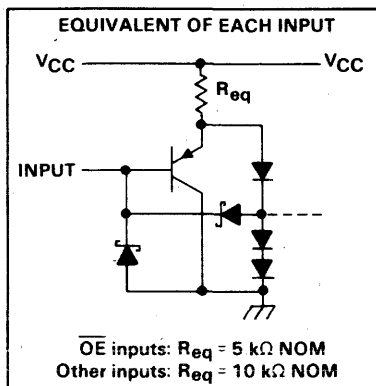
The SN54PL16' is characterized for operation over the full military temperature range of - 55°C to 125°C. The SN74PL16' is characterized for operation from 0°C to 70°C.

## pin assignments in operating mode (voltages at pins 1 and 11 less than $V_{IH}$ )

SN54PL' . . . J PACKAGE  
SN74PL' . . . N PACKAGE  
(TOP VIEWS)



## schematics of inputs and outputs



## ADVANCE INFORMATION

This document contains information on a new product. Specifications are subject to change without notice.

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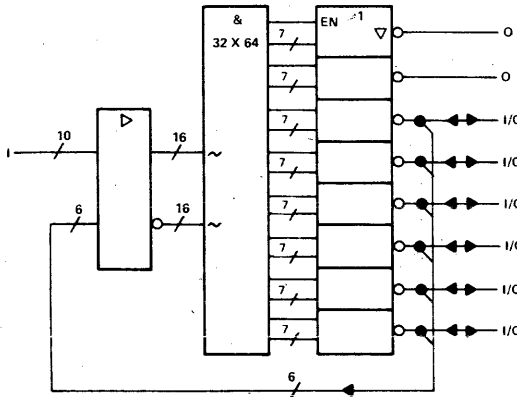
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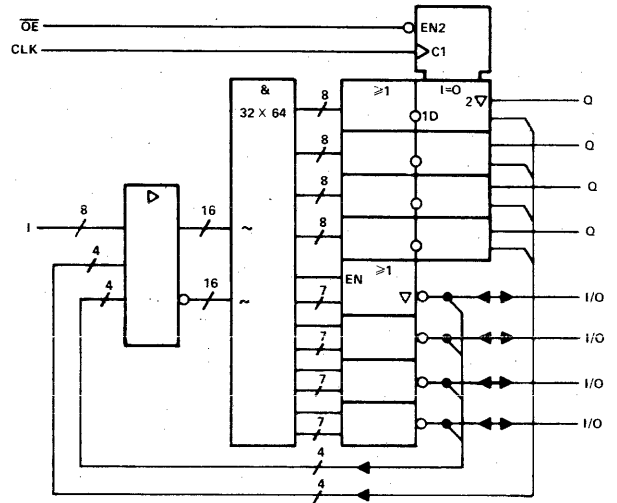
TYPES SN54PL16L8, SN54PL16R4, SN54PL16R6, SN54PL16R8,  
SN74PL16L8, SN74PL16R4, SN74PL16R6, SN74PL16R8  
FIXED-OR ARRAYS

functional block diagrams (positive logic)

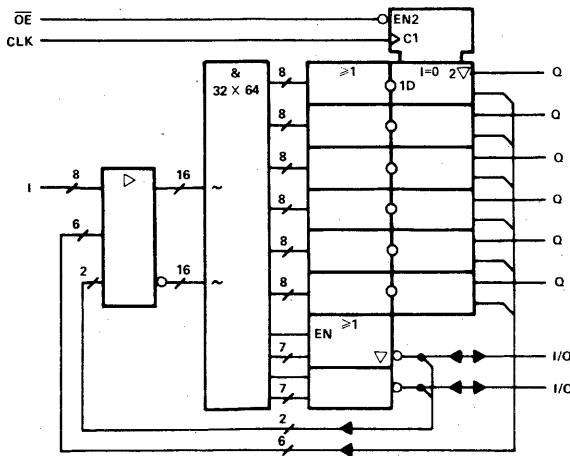
'PL16L8



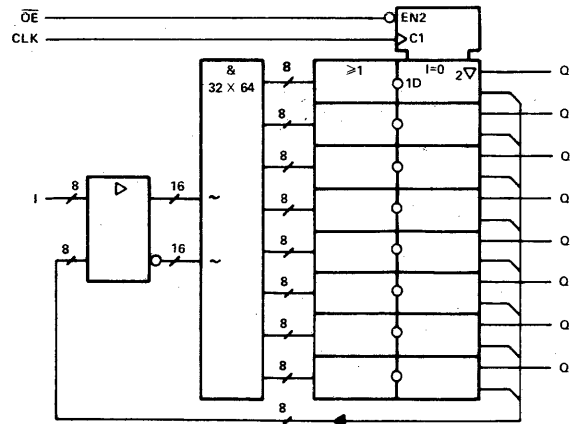
'PL16R4



'PL16R6



'PL16R8



~ denotes fused inputs

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

**TYPES SN54PL16L8, SN54PL16R4, SN54PL16R6, SN54PL16R8,  
SN74PL16L8, SN74PL16R4, SN74PL16R6, SN74PL16R8  
FIXED-OR ARRAYS**

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage (see Note 1) .....	5.5 V
Voltage applied to a disabled output (see Note 1) .....	5.5 V
Operating free-air temperature range: SN54PL <sup>1</sup> .....	-55°C to 125°C
SN74PL <sup>1</sup> .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C

NOTE 1: These ratings apply except for programming pins during a programming cycle.

**recommended operating conditions**

PARAMETER	SN54PL16 <sup>1</sup>			SN74PL16 <sup>1</sup>			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
$V_{CC}$ Supply voltage	4.5	5	5.5	4.75	5	5.25	V
$V_{IH}$ High-level input voltage	2		5.5	2		5.5	V
$V_{IL}$ Low-level input voltage			0.8			0.8	V
$I_{OH}$ High-level output current			-2			-3.2	mA
$I_{OL}$ Low-level output current			12			24	mA
$T_A$ Operating free-air temperature	-55		125	0		70	°C

**electrical characteristics, over recommended free-operating temperature range**

PARAMETER	TEST CONDITIONS <sup>†</sup>	SN54PL16 <sup>1</sup>			SN74PL16 <sup>1</sup>			UNIT
		MIN	TYP <sup>‡</sup>	MAX	MIN	TYP <sup>‡</sup>	MAX	
$V_{IK}$	$V_{CC} = \text{MIN}, I_I = -18 \text{ mA}$			-1.5			-1.5	V
$V_{OH}$	$V_{CC} = \text{MIN}, I_{OH} = \text{MAX}$	2.4	3.2		2.4	3.3		V
$V_{OL}$	$V_{CC} = \text{MIN}, I_{OL} = \text{MAX}$		0.25	0.4		0.35	0.5	V
$I_{OZH}$	Outputs	$V_{CC} = \text{MAX}, V_O = 2.7 \text{ V}$		20		20		$\mu\text{A}$
	I/O ports							
$I_{OZL}$	Outputs	$V_{CC} = \text{MAX}, V_O = 0.4 \text{ V}$		-20		-20		$\mu\text{A}$
	I/O ports							
$I_I$	$V_{CC} = \text{MAX}, V_I = 5.5 \text{ V}$			0.1			0.1	mA
$I_{IH}$	$V_{CC} = \text{MAX}, V_I = 2.7 \text{ V}$			20			20	$\mu\text{A}$
$I_{IL}$	OE Input	$V_{CC} = \text{MAX}, V_I = 0.4 \text{ V}$		-1 Parts		-0.4		mA
	All others							
	OE Input			-2 Parts		-0.2		
	All others							
$I_O^{\S}$	$V_{CC} = \text{MAX}, V_O = 2.25 \text{ V}$	-30	-125		-30	-125	mA	
$I_{CC}$	$V_{CC} = \text{MAX}, V_I = 0 \text{ V}, \text{OE at } 4.5 \text{ V}, \text{Outputs open}$	-1 Parts		140	185	140	180	mA
		-2 Parts		75	95	70	90	

<sup>†</sup>For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

<sup>‡</sup>All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}$ .

<sup>§</sup>The output conditions have been chosen to produce a current that closely approximates one half of the true short-circuit output current,  $I_{OS}$ .

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**TYPES SN54PL16L8, SN54PL16R4, SN54PL16R6, SN54PL16R8,  
SN74PL16L8, SN74PL16R4, SN74PL16R6, SN74PL16R8  
FIXED-OR ARRAYS**

**'PL16R4, 'PL16R6, 'PL16R8 timing requirements**

		-1 PARTS			-2 PARTS			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$f_{\text{clock}}$	Clock frequency	0		35				MHz
$t_w$	Pulse duration, clock high or low	12						ns
$t_{\text{su}}$	Setup time, input or feedback before CLK↑	15						ns
$t_h$	Hold time, input or feedback after CLK↑	0						ns

**switching characteristics over recommended supply voltage and operating free-air temperature ranges (unless otherwise noted)**

PARAMETER	FROM	TO	TEST CONDITIONS	-1 PARTS			-2 PARTS			UNIT	
				MIN	TYP†	MAX	MIN	TYP†	MAX		
$f_{\text{max}}$			$R_L = 500 \Omega$ $C_L = 50 \text{ pf}$	35	45		35			MHz	
$t_{\text{pd}}$	I, I/O	O, I/O			16	25		32			ns
$t_{\text{pd}}$	CLK↑	Q		7	10	15		20			ns
$t_{\text{en}}$	$\overline{\text{OE}}\downarrow$	Q			18	25		36			ns
$t_{\text{dis}}$	$\overline{\text{OE}}\uparrow$	Q			10	15		20			ns
$t_{\text{en}}$	I, I/O	O, I/O			18	25		36			ns
$t_{\text{dis}}$	I, I/O	O, I/O			14	25		28			ns

†All typical values are at  $V_{\text{CC}} = 5 \text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

**programming parameters,  $T_A = 25^\circ\text{C}$**

		MIN	NOM	MAX	UNIT
$V_{\text{CC}}$	Verify-level supply voltage	4.5	5.0	5.5	V
$V_{\text{IH}}$	High-level input voltage	2		5.5	V
$V_{\text{IL}}$	Low-level input voltage			0.8	V
$V_{\text{IHH}}$	Program-pulse input voltage	10.25	10.5	10.75	V
$I_{\text{IHH}}$	Program-pulse input current	PO	20	50	mA
		PBM ENABLE, L/R	10	25	
		PI, PA	1.5	5	
		$V_{\text{CC}}$	250	400	
$t_{\text{w1}}$	Program-pulse duration at PO pins	10		50	$\mu\text{s}$
$t_{\text{w2}}$	Pulse duration at PGM VERIFY	100			ns
	Program-pulse duty cycle at PO pins			25	%
$t_{\text{su}}$	Setup time	100			ns
$t_h$	Hold time	100			ns
$t_{\text{d1}}$	Delay time from $V_{\text{CC}}$ to 5 V to PGM VERIFY 1	100			$\mu\text{s}$
$t_{\text{d2}}$	Delay time from PGM VERIFY 1 to valid output	200			ns
	Input voltage at pins 1 and 11 to open verify-protect (security) fuse	20	21	22	V
	Input current to open verify-protect (security) fuse			400	mA
	Pulse duration to open verify-protect (security) fuse	20		50	ms

**FIELD-PROGRAMMABLE LOGIC**

**TYPES SN54PL20L8, SN54PL20R4, SN54PL20R6, SN54PL20R8  
SN74PL20L8, SN74PL20R4, SN74PL20R6, SN74PL20R8  
FIXED-OR ARRAYS**

D2706, DECEMBER 1982—REVISED SEPTEMBER 1983

- Standard 24-Pin, 300-mil Packages
- Output Registers Have Preload Capability
- Output Registers Automatically Clear During Power-Up
- Choice of Operating Speeds
  - 1 Parts . . . 30 MHz Max, Standard power
  - 2 Parts . . . 20 MHz Max, Half power

DEVICE	I INPUTS	3-STATE 0 OUTPUTS	REGISTERED Q OUTPUTS	I/O PORTS
'PL20L8	14	2	0	6
'PL20R4	12	0	4 (3-state buffers)	4
'PL20R6	12	0	6 (3-state buffers)	2
'PL20R8	12	0	8 (3-state buffers)	0

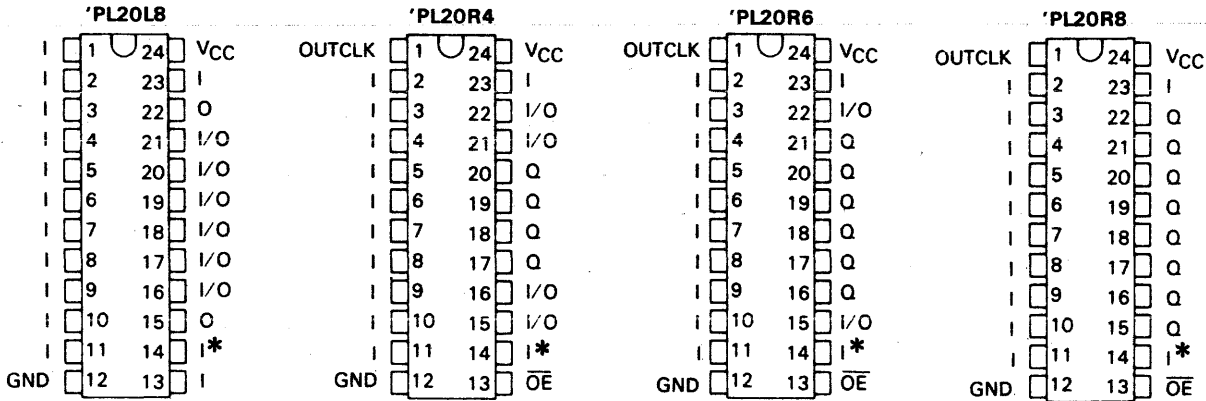
**description**

These fixed-OR arrays provide 3-state outputs for bus-oriented systems. The 'PL20L8, 'PL20R4, and 'PL20R6 have output registers that can be loaded from the I/O pins by a preload procedure. All the outputs are automatically set to a low level when power is applied. The -1 and -2 parts offer a choice of operating frequency, switching times, and power dissipation.

The SN54PL20' is characterized for operation over the full military temperature range of -55°C to 125°C. The SN74PL20' is characterized for operation from 0°C to 70°C.

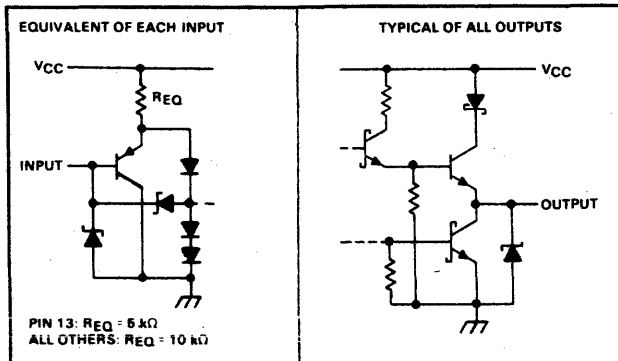
**pin assignments in operating mode (voltages at pins 1 and 13 less than  $V_{IH}$ )**

SN54' ... JT PACKAGE  
SN74' ... NT PACKAGE  
(TOP VIEW)



\*Pin 14 is also used for the preload procedure on the last page of this data sheet.

**schematics of inputs and outputs**



**ADVANCE INFORMATION**

This document contains information on a new product. Specifications are subject to change without notice.

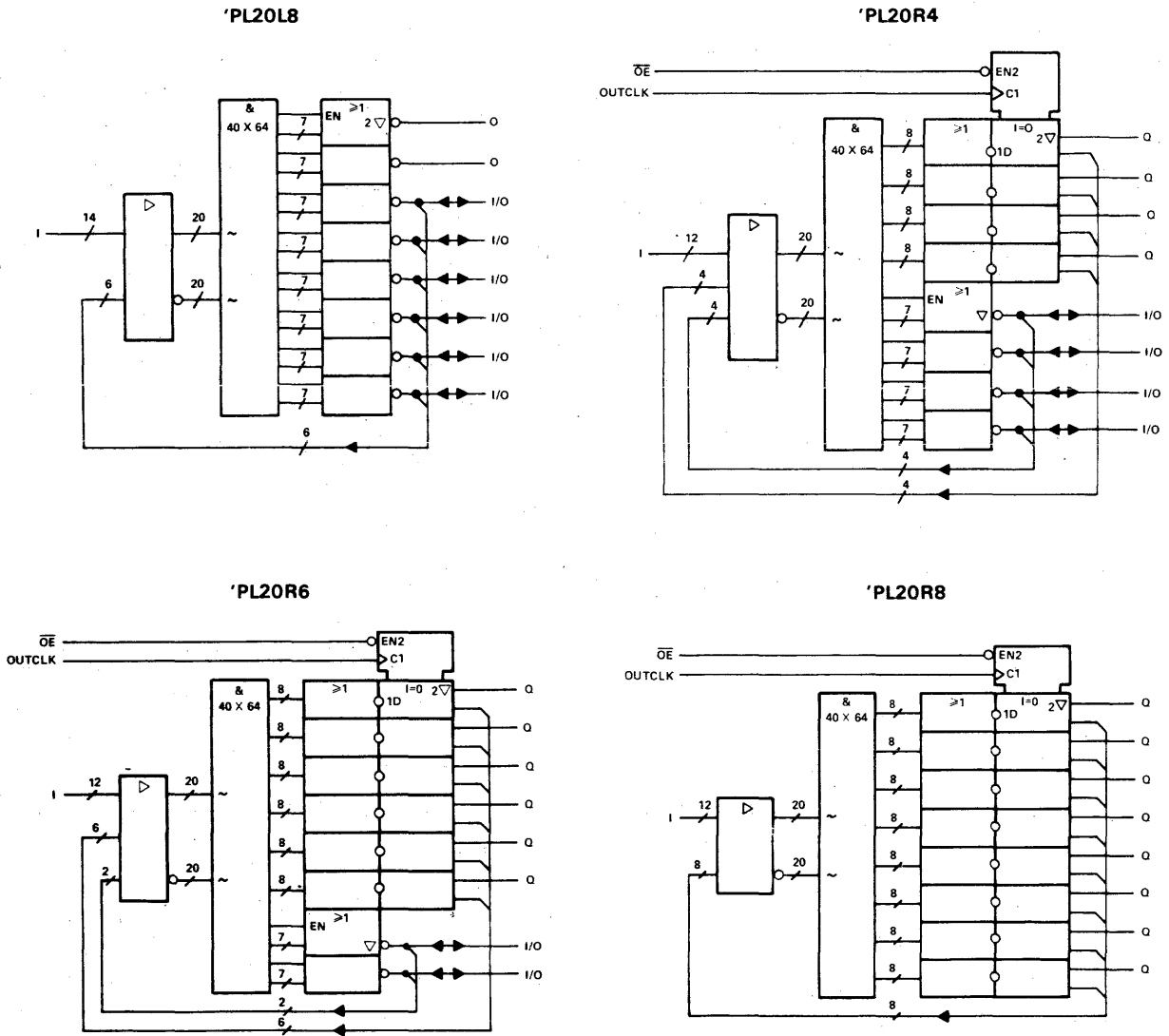
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**TEXAS INSTRUMENTS**

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# TYPES SN54PL20L8, SN54PL20R4, SN54PL20R6, SN54PL20R8 SN74PL20L8, SN74PL20R4, SN74PL20R6, SN74PL20R8 FIXED-OR ARRAYS

functional block diagrams (positive logic)



**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage (see Note 1) .....	5.5 V
Voltage applied to a disabled output (see Note 1) .....	5.5 V
Operating free-air temperature range: SN54PL' .....	-55°C to 125°C
SN74PL' .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C

NOTE 1: These ratings apply except for programming pins during a programming cycle or during preload cycle.

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

**TEXAS  
INSTRUMENTS**

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**TYPES SN54PL20L8, SN54PL20R4, SN54PL20R6, SN54PL20R8,  
SN74PL20L8, SN74PL20R4, SN74PL20R6, SN74PL20R8  
FIXED-OR ARRAYS**

**recommended operating conditions**

		SN54PL19R'			SN74PL19R'			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
V <sub>IH</sub>	High-level input voltage	2		5.5	2		5.5	V
V <sub>IL</sub>	Low-level input voltage			0.8			0.8	V
I <sub>OH</sub>	High-level output current			-2			-3.2	mA
I <sub>OL</sub>	Low-level output current			12			24	mA
T <sub>A</sub>	Operating free-air temperature	-55		125	0		70	°C

**electrical characteristics over recommended free-air operating temperature range**

PARAMETER	TEST CONDITIONS†	SN54PL20'			SN74PL20'			UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>IK</sub>	V <sub>CC</sub> = MIN, I <sub>I</sub> = -18 mA			-1.5			-1.5	V
V <sub>OH</sub>	V <sub>CC</sub> = MIN, I <sub>OH</sub> = MAX	2.4	3.2		2.4	3.3		V
V <sub>OL</sub>	V <sub>CC</sub> = MIN, I <sub>OL</sub> = MAX		0.25	0.4		0.35	0.5	V
I <sub>OZH</sub>	O, Q outputs	V <sub>CC</sub> = MAX, V <sub>IH</sub> = 2.7 V		20	20			μA
	I/O ports			100				
I <sub>OZL</sub>	O, Q outputs	V <sub>CC</sub> = MAX, V <sub>IH</sub> = 0.4 V		-20	-20			μA
	I/O ports			-250				
I <sub>I</sub>	OE Input	V <sub>CC</sub> = MAX, V <sub>I</sub> = 5.5 V		0.2	0.2			mA
	All others			0.1				
I <sub>IH</sub>	OE Input	V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.7 V		40	40			μA
	All others			20				
I <sub>IL</sub>	OE Input	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V		-0.4	-0.4			mA
	All others			-0.2				
I <sub>O</sub> <sup>§</sup>	V <sub>CC</sub> = MAX, V <sub>O</sub> = 2.25 V	-30		-125	-30		-125	mA
I <sub>CC</sub>	-1 Parts	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0 V, Outputs open, OE at V <sub>IH</sub>		150	200	150	200	mA
	-2 Parts			75	100	75	100	

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡All typical values are V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

§The output conditions have been chosen to produce a current that closely approximates one half the true short-circuit current, I<sub>OS</sub>.

**'PL20R4, 'PL20R6, 'PL20R8 timing requirements**

		-1 PARTS		-2 PARTS		UNIT
		MIN	MAX	MIN	MAX	
f <sub>clock</sub>	Clock frequency	0	30	0	20	MHz
t <sub>w</sub>	Pulse duration, clock high or low	12		12		ns
t <sub>su</sub>	Setup time, input or feedback before OUTCLK†	15		15		ns
t <sub>h</sub>	Hold time, input or feedback before OUTCLK†	0		0		ns

**switching characteristics over recommended operating free-air temperature range (unless otherwise noted)**

PARAMETER	FROM	TO	TEST CONDITIONS	-1 PARTS			-2 PARTS			UNIT
				MIN	TYP‡	MAX	MIN	TYP‡	MAX	
f <sub>max</sub>			R <sub>L</sub> = 500 Ω, C <sub>L</sub> = 50 pF	30			20			MHz
t <sub>pd</sub>	I, I/O	O, I/O		16			25			ns
t <sub>pd</sub>	OUTCLK †	Q		12			20			ns
t <sub>en</sub>	OE	Q		8			15			ns
t <sub>dis</sub>	OE†	Q		6			12			ns
t <sub>en</sub>	I, I/O	O, I/O		18			25			ns
t <sub>dis</sub>	I, I/O	O, I/O		13			20			ns

‡All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

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INSTRUMENTS**

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TYPES SN54PL20L8, SN54PL20R4, SN54PL20R6, SN54PL20R8,  
SN74PL20L8, SN74PL20R4, SN74PL20R6, SN74PL20R8  
FIXED-OR ARRAYS

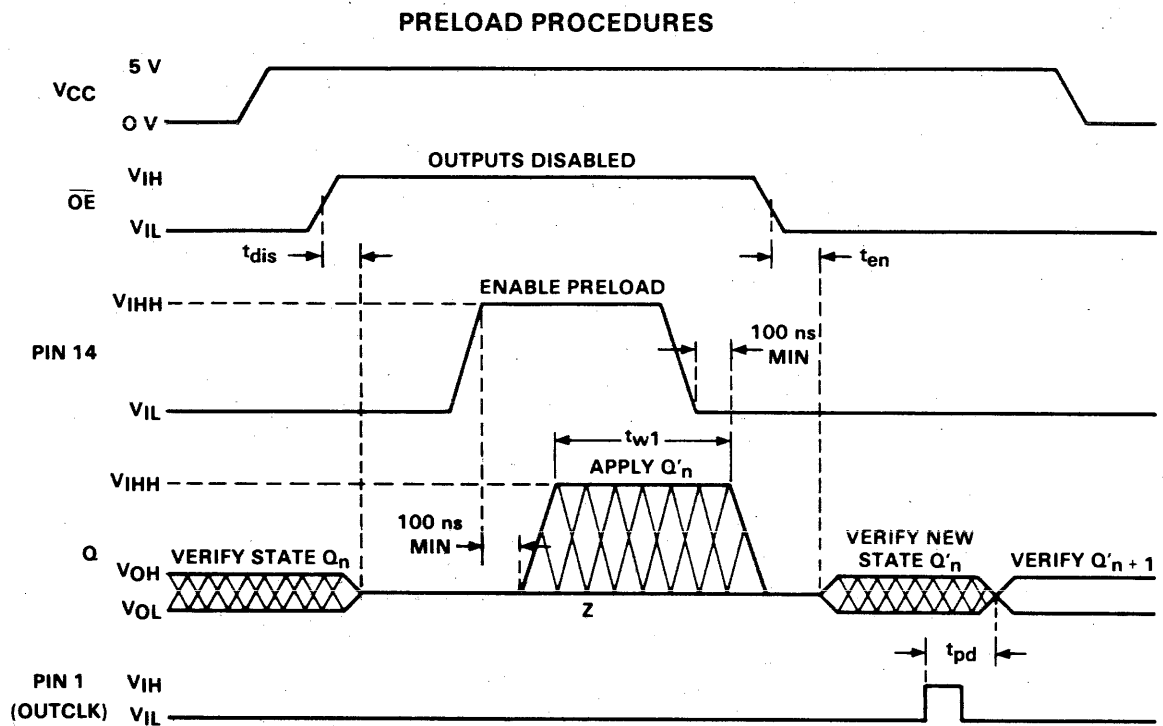


FIGURE 1—PRELOAD WAVEFORMS

**preload procedure for registered outputs**

- Step 1 Pin 13 to V<sub>IH</sub>, Pin 1 to V<sub>IL</sub>, and V<sub>CC</sub> to 5 volts.
- Step 2 Pin 14 to V<sub>IHH</sub> for 10 to 50 microseconds.
- Step 3 Apply an open circuit for a low and V<sub>IHH</sub> for a high at the Q outputs.
- Step 4 Pin 14 to V<sub>IL</sub>.
- Step 5 Remove the voltages applied to the outputs.
- Step 6 Pin 13 to V<sub>IL</sub>.
- Step 7 Check the output states to verify preload.



# FIELD-PROGRAMMABLE LOGIC

## TYPES SN54PL839, SN54PL840, SN74PL839, SN74PL840 14 X 32 X 6 FIELD-PROGRAMMABLE LOGIC ARRAYS

D2708, DECEMBER 1982—REVISED SEPTEMBER 1983

- Input-to-Output Propagation Delay . . . 10 ns Typical
- 24-Pin, 300-mil Slim Line Packages
- Power Dissipation . . . 650 mW Typical
- Programmable Output Polarity

### LOGIC FUNCTION

$$f(i) = P_0 + P_1 \dots P_{31} \text{ for polarity link intact}$$

$$f(i) = \overline{P_0} * \overline{P_1} * \dots * \overline{P_{31}} \text{ for polarity link open}$$

where P<sub>0</sub> through P<sub>31</sub> are product terms

### description

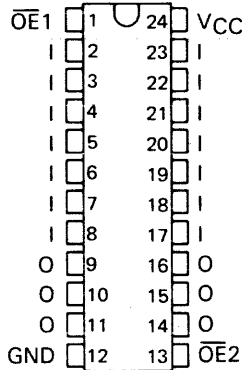
The 'PL839 (3-state outputs and the 'PL840 (open-collector outputs) are TTL field-programmable logic arrays containing 32 product terms (AND terms) and six sum terms (OR terms). Each of the six sum-of-products output functions can be programmed either high or low true. The true condition of each output function is activated by the programmed logical minterms of 14 input variables. The outputs are controlled by two chip-enable pins to allow output inhibit and expansion of terms.

These devices provide high-speed data-path logic replacement where several conventional SSI functions can be designed into a single package.

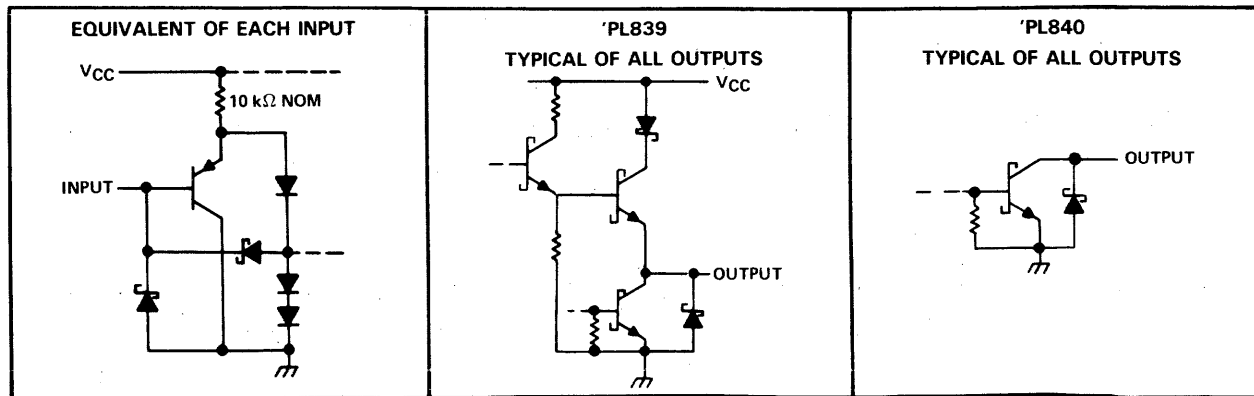
The SN54PL839 and SN54PL840 are characterized for operation over the full military temperature range of -55°C to 125°C. The NS74PL839 and SN74PL840 are characterized for operation from 0°C to 70°C.

### pin assignments in operating mode (pin 1 is less than V<sub>IHH</sub>)

SN54PL' . . . JT PACKAGE  
SN74PL' . . . NT PACKAGE  
(TOP VIEW)



### schematics of inputs and outputs



### ADVANCE INFORMATION

This document contains information on a new product. Specifications are subject to change without notice.

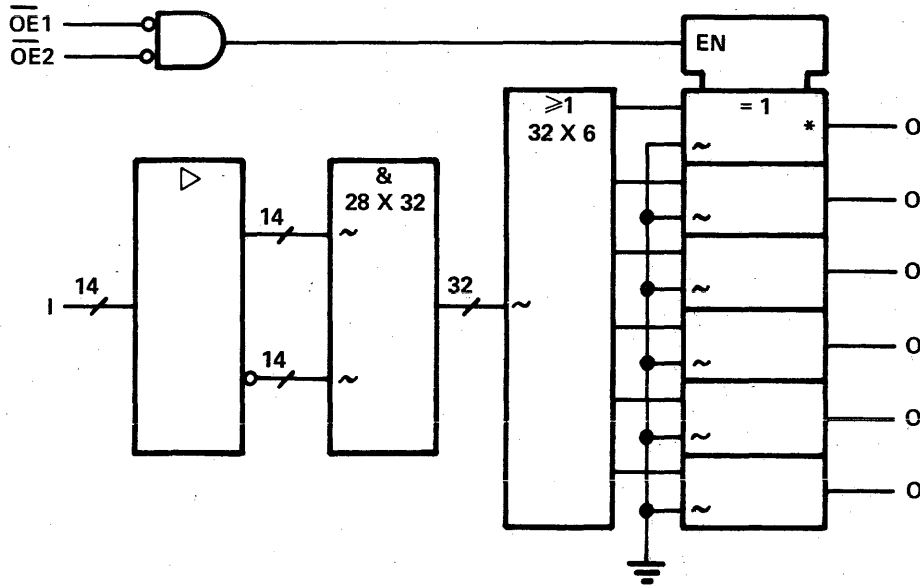
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**TYPES SN54PL839, SN54PL840, SN74PL839, SN74PL840**  
**14 X 32 X 6 FIELD-PROGRAMMABLE LOGIC ARRAYS**

functional block diagram (positive logic)



~ denotes fused inputs.

\* PL839 has 3-state ( $\nabla$ ) outputs; PL840 has open-collector ( $\square$ ) outputs.

**absolute maximum ratings**

Supply Voltage, $V_{CC}$ .....	7 V
Input Voltage .....	5.5 V
Off-State Output Voltage .....	5.5 V
Operating Free-air Temperature Range SN54PL839, SN54PL840 .....	-55°C to 125°C
SN74PL839, SN74PL840 .....	0°C to 70°C
Storage Temperature .....	-65°C to 150°C

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

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**TYPES SN54PL839, SN54PL840, SN74PL839, SN74PL840**  
**14 X 32 X 6 FIELD-PROGRAMMABLE LOGIC ARRAYS**

**recommended operating conditions**

	SN54PL'			SN74PL'			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC}$	4.5	5	5.5	4.75	5	5.25	V
High-level input voltage, $V_{IH}$	2			2			V
Low-level input voltage, $V_{IL}$			0.8			0.8	V
High-level output current, $I_{OH}$			-2			-3.2	mA
Low-level output current, $I_{OL}$			12			24	mA
Operating free-air temperature, $T_A$	-55		125	0		70	°C

**electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)**

PARAMETER	TEST CONDITIONS†	SN54PL'			SN74PL'			UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
$V_{IK}$	$V_{CC} = \text{MIN}, I_I = -18 \text{ mA}$			-1.5			-1.5	V
$V_{OH}$	$V_{CC} = \text{MIN}, I_{OH} = \text{MAX}$	2.4	3.2		2.4	3		V
$V_{OL}$	$V_{CC} = \text{MIN}, I_{OL} = \text{MAX}$		0.25	0.5		0.37	0.5	V
$I_I$	$V_{CC} = \text{MAX}, V_I = 5.5 \text{ V}$			0.1			0.1	mA
$I_{IH}$	$V_{CC} = \text{MAX}, V_I = 2.7 \text{ V}$			20			20	µA
$I_{IL}$	$V_{CC} = \text{MAX}, V_I = 0.4 \text{ V}$			-0.4			-0.4	mA
$I_O^{\S}$	$V_{CC} = \text{MAX}, V_O = 2.25 \text{ V}$	-30	-50	-112	-30	-50	-112	mA
$I_{OZH}$	$V_{CC} = \text{MAX}, V_O = 2.7 \text{ V}$			20			20	µA
$I_{OZL}$	$V_{CC} = \text{MAX}, V_O = 0.4 \text{ V}$			-20			-20	µA
$I_{CC}$	$V_{CC} = \text{MAX},$ $\overline{OE}$ inputs at $V_{IH}$		130	200		130	190	mA

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}$ .

§The output conditions have been chosen to produce a current that closely approximates one half of the true short-circuit current,  $I_{OS}$ .

**switching characteristics over recommended operating free-air temperature range (unless otherwise noted)**

PARAMETER	FROM	TO	TEST CONDITIONS	SN54PL'			SN74PL'			UNIT
				MIN	TYP‡	MAX	MIN	TYP‡	MAX	
$t_{pd}$	Input	Output	$R_L = 500 \text{ to GND},$ $C_L = 50 \text{ pF to GND}$		10	25		10	20	ns
$t_{en}$	Pin 1	Output	$R_{L1} = 500 \text{ to } 7 \text{ V},$ $R_{L2} = 500 \text{ to GND},$ $C_L = 50 \text{ pF to GND}$		9	16		9	13	ns
$t_{dis}$	Pin 13				8	15		8	12	

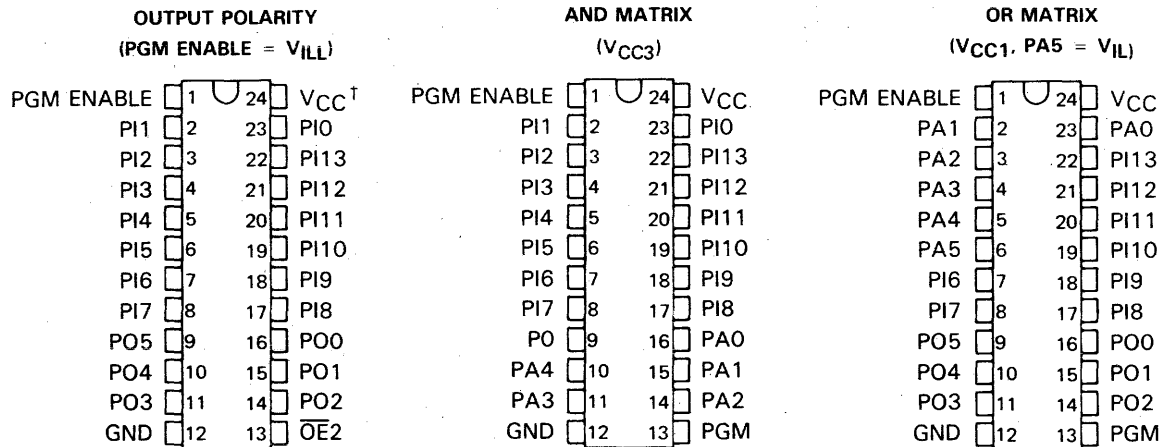
‡All typical values are at  $V_{CC} = 5 \text{ V}, T_A = 25^\circ\text{C}$ .

**TEXAS  
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# TYPES SN54PL839, SN54PL840, SN74PL839, SN74PL840 14 X 32 X 6 FIELD-PROGRAMMABLE LOGIC ARRAYS

pin assignment in programming mode (pin 1  $\geq$   $V_{IH}$ ) top views



$\dagger V_{CC} = V_{CC2}$  for program and  $V_{CC1}$  for verify

programming parameters,  $T_A = 25^\circ\text{C}$

PARAMETER		MEASURED AT	PROGRAMMING MODE	MIN	TYP	MAX	UNIT
$V_{IH}$	Program high-level input voltage	PGM ENABLE	AND, OR	16.5	17	17.5	V
		PO pins	Polarity				
$V_{ILL}$	Program low-level input voltage	PGM ENABLE	Any	0		0.4	V
$I_{IH}$	Program-level input current	PO pins	Polarity	100			mA
		PGM ENABLE	AND, OR	150			
$V_{IX}$	Program-level input voltage	PO0 thru PO5	Polarity	9.5	10	10.5	V
		PGM	AND, OR				
$I_{IX}$	Program-level input current	PI pins	AND	0.6		2	mA
		$\overline{OE}2$	Polarity			5	
		PO0 thru PO5	OR			5	
$V_{CC1}$	Programming supply voltage	$V_{CC}$	OR	8.5	8.75	9	V
$I_{CC1}$	Programming supply current	$V_{CC}$	OR		250	400	mA
$V_{CC2}$	Programming supply voltage	$V_{CC}$	Polarity		0	0.4	V
$V_{CC3}$	Programming supply voltage	$V_{CC}$	AND	4.75	5	5.25	V
$V_{IH}$	High-level input voltage	Any	Any	2			V
$V_{IL}$	Low-level input voltage	Any	Any	0		0.8	V
$V_{OH}$	High-level output voltage	Any	Any	2.4	3.2		V
$V_{OL}$	Low-level output voltage	Any	Any	0.25		0.5	V
$t_w$	Program pulse duration	PO0 thru PO5	Polarity	50		1000	$\mu\text{s}$
		PGM	AND, OR				
	Program pulse duty cycle	PO0 thru PO5	Polarity	10		50	%
		PGM	AND, OR				
$t_d$	Delay time	Any	Any	10		$\mu\text{s}$	
$t_r$	Rise time	Any	Any	25		$\mu\text{s}$	

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**FIELD-PROGRAMMABLE LOGIC**

**TYPES SN54PLR19L8, SN54PLR19R4, SN54PLR19R6, SN54PLR19R8  
SN74PLR19L8, SN74PLR19R4, SN74PLR19R6, SN74PLR19R8  
REGISTERED-INPUT FIXED-OR ARRAYS**

D2709, DECEMBER 1982—REVISED SEPTEMBER 1983

- Standard 24-Pin, 300-mil Packages
- Output Registers Have Preload Capability
- Output Registers Automatically Clear During Power-Up
- Choice of Operating Speeds
  - 1 Parts . . . 30 MHz Max, Standard power
  - 2 Parts . . . 20 MHz Max, Half power

DEVICE	I/D INPUTS	I INPUTS	3-STATE O OUTPUTS	REGISTERED Q OUTPUTS	I/O PORTS
'PLR19L8	11	2	2	0	6
'PLR19R4	11	0	0	4 (3-state buffers)	4
'PLR19R6	11	0	0	6 (3-state buffers)	2
'PLR19R4	11	0	0	8 (3-state buffers)	0

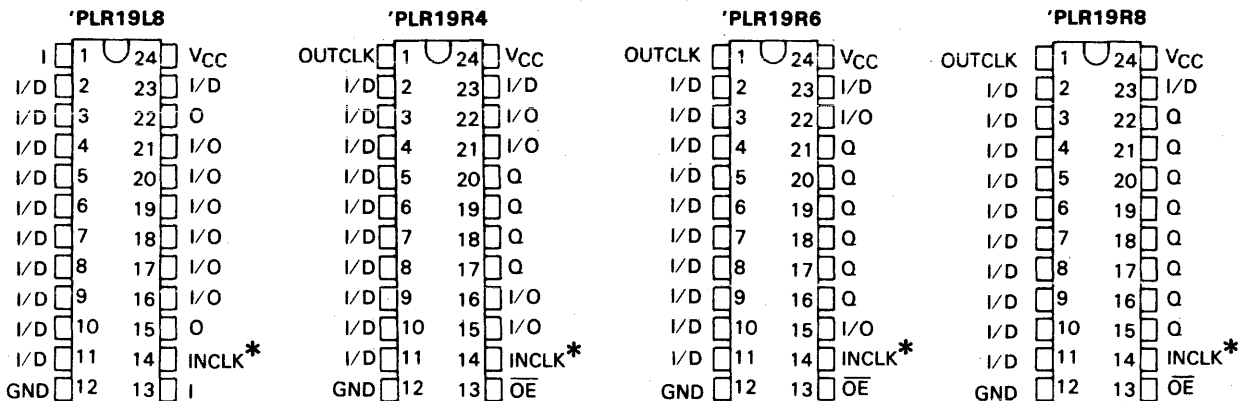
**description**

These fixed-OR arrays with eleven data inputs feature input registers that can be used as they are or be programmed into buffers. Some outputs of the 'PL19R8, 'PL19R6, and 'PL19R4 have registers that can be loaded from the I/O pins by a preload procedure, while others are I/O ports and standard 3-state outputs. All the outputs are automatically set to a low level when power is applied. The -1 and -2 parts offer a choice of operating frequency, switching times, and power dissipation.

The SN54PLR19' is characterized for operation over the full military temperature range of -55°C to 125°C. The SN74PLR19' is characterized for operation from 0°C to 70°C.

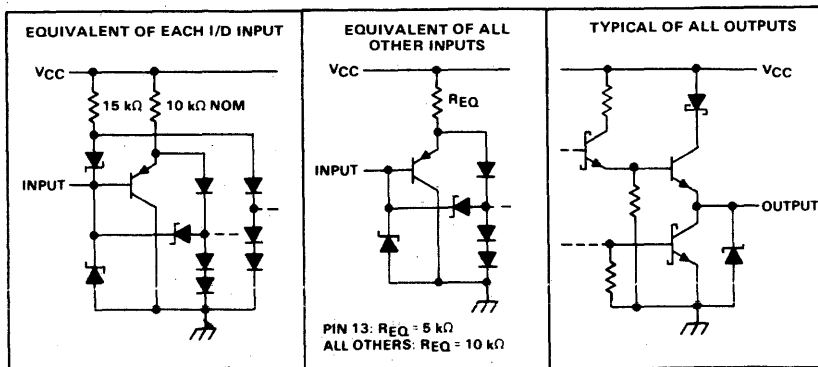
**pin assignments in operating mode (voltages at pins 1 and 13 less than  $V_{IH}$ )**

SN54' ... JT PACKAGE  
SN74' ... NT PACKAGE  
(TOP VIEW)



\*Pin 14 is also used for the preload procedure on the last page of this data sheet.

**schematics of inputs and outputs**



**ADVANCE INFORMATION**

This document contains information on a new product. Specifications are subject to change without notice.

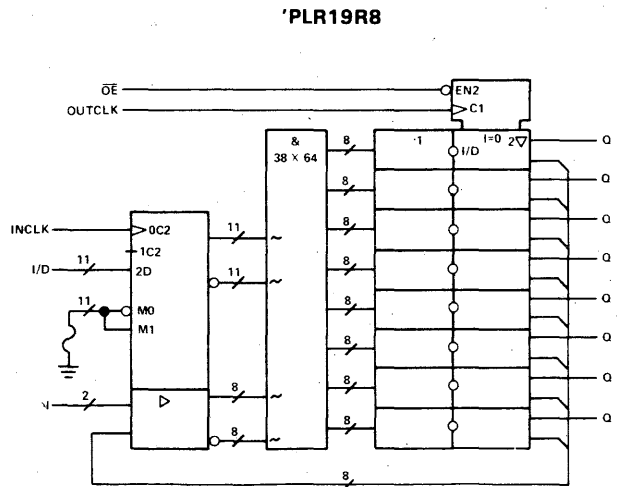
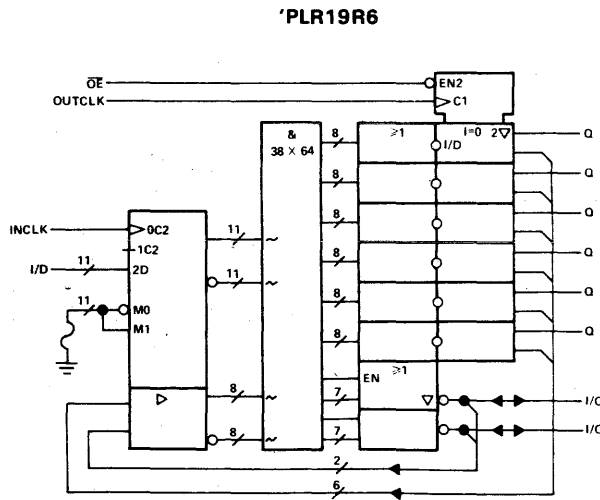
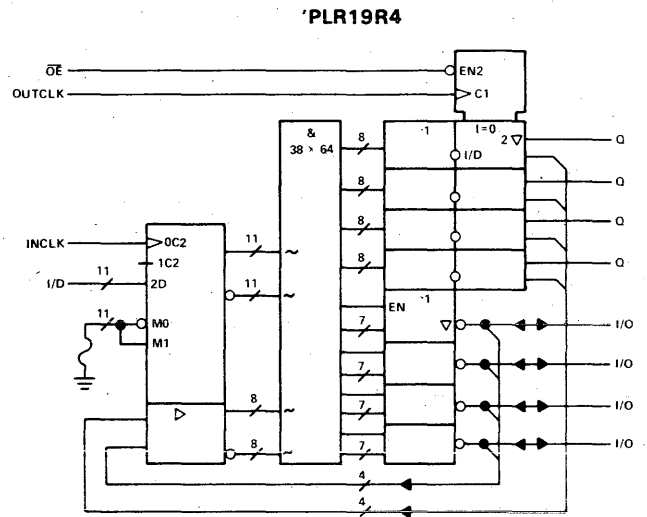
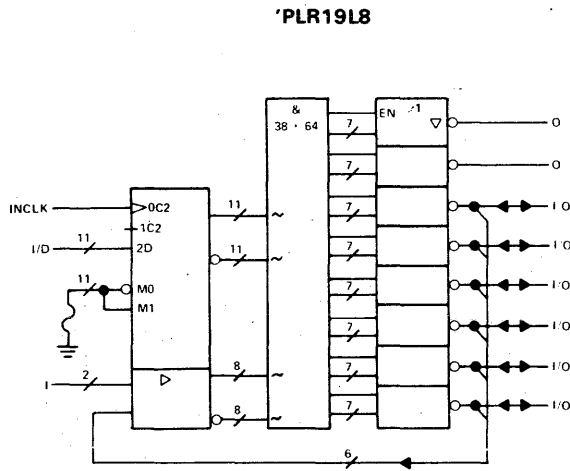
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**TYPES SN54PLR19L8, SN54PLR19R4, SN54PLR19R6, SN54PLR19R8  
SN74PLR19L8, SN74PLR19R4, SN74PLR19R6, SN74PLR19R8  
REGISTERED-INPUT FIXED-OR ARRAYS**

functional block diagrams (positive logic)



**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage (see Note 1) .....	5.5 V
Voltage applied to a disabled output (see Note 1) .....	5.5 V
Operating free-air temperature range: SN54PLR' .....	-55°C to 125°C
SN74PLR' .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C

NOTE 1: These ratings apply except for programming pins during a programming cycle or during a preload cycle.

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

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**TEXAS  
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**TYPES SN54PLR19L8, SN54PLR19R4, SN54PLR19R6, SN54PLR19R8,  
SN74PLR19L8, SN74PLR19R4, SN74PLR19R6, SN74PLR19R8  
REGISTERED-INPUT FIXED-OR ARRAYS**

**recommended operating conditions**

		SN54PLR19'			SN74PLR19'			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
V <sub>IH</sub>	High-level input voltage	2		5.5	2		5.5	V
V <sub>IL</sub>	Low-level input voltage			0.8			0.8	V
I <sub>OH</sub>	High-level output current			-2			-3.2	mA
I <sub>OL</sub>	Low-level output current			12			24	mA
T <sub>A</sub>	Operating free-air temperature	-55		125	0		70	°C

**electrical characteristics over recommended free-air operating temperature range**

PARAMETER	TEST CONDITIONS†	SN54PLR19'			SN74PLR19'			UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>IK</sub>	V <sub>CC</sub> = MIN, I <sub>I</sub> = -18 mA			-1.5			-1.5	V
V <sub>OH</sub>	V <sub>CC</sub> = MIN, I <sub>OH</sub> = MAX	2.4	3.2		2.4	3.3		V
V <sub>OL</sub>	V <sub>CC</sub> = MIN, I <sub>OL</sub> = MAX		0.25	0.4		0.35	0.5	V
I <sub>OZH</sub>	Outputs			20			20	μA
	I/O ports			100			100	
I <sub>OZL</sub>	Outputs			-20			-20	μA
	I/O ports			-250			-250	
I <sub>I</sub>	OE Input			0.2			0.2	mA
	I/D Inputs			0.1			0.1	
	All others			0.1			0.1	
I <sub>IH</sub>	OE Input			40			40	μA
	I/D Inputs			20			0.1	
	All others			20			0.1	
I <sub>IL</sub>	OE Input			-0.4			-0.4	mA
	I/D Inputs			-0.6			-0.6	
	All others			-0.2			-0.2	
I <sub>O</sub> §	V <sub>CC</sub> = MAX, V <sub>O</sub> = 2.25 V	-30		-125	-30		-125	mA
I <sub>CC</sub>	-1 Parts		150	200		150	200	mA
	-2 Parts	Outputs open	75	100	75	100		

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡All typical values are V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

§The output conditions have been chosen to produce a current that closely approximates one half the true short-circuit current, I<sub>OS</sub>.

**input timing requirements**

		MIN	MAX	UNIT
f <sub>clock</sub>	Clock frequency	0	40	MHz
t <sub>w</sub>	Clock pulse duration, clock high or low	10		ns
t <sub>su</sub>	Setup time, I/D input before INCLK!	12		ns
t <sub>h</sub>	Hold time, I/D input before INCLK!	0		ns

**'PLR19R4, 'PLR19R6, 'PLR19R8 timing requirements**

		MIN	MAX	UNIT
f <sub>clock</sub>	Clock frequency	0	30	MHz
t <sub>w</sub>	Clock pulse duration, clock high or low	12		ns
t <sub>su</sub>	Setup time, input or feedback before OUTCLK!	15		ns
t <sub>h</sub>	Hold time, input or feedback before OUTCLK!	0		ns

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**TYPES SN54PLR19L8, SN54PLR19R4, SN54PLR19R6, SN54PLR19R8,  
SN74PLR19L8, SN74PLR19R4, SN74PLR19R6, SN74PLR19R8  
REGISTERED-INPUT FIXED-OR ARRAYS**

switching characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	FROM	TO	INPUT MODE	TEST CONDITIONS	- 1 PARTS			- 2 PARTS			UNIT
					MIN	TYP†	MAX	MIN	TYP†	MAX	
$f_{max}$			Either	$R_L = 500\Omega,$ $C_L = 50\text{ pF}$	30			20			MHz
$t_{pd}$	I, I/O	I/O, O	Either			16			25		ns
$t_{pd}$	OUTCLK ↑	Q	Either			12			20		ns
$t_{en}$	$\overline{OE}$ ↓	Q	Either			8			15		ns
$t_{dis}$	$\overline{OE}$ ↑	Q	Either			6			12		ns
$t_{pd}$	INCLK ↑	I/O, O	Registered			23			32		ns
$t_{en}$	INCLK ↑	I/O, O, Q	Registered			25			35		ns
$t_{dis}$	INCLK ↑	I/O, O, Q	Registered			20			30		ns
$t_{pd}$	I/D	I/O, O	Buffered			20			30		ns
$t_{en}$	I/D, I/O	I/O	Buffered			22			32		ns
$t_{dis}$	I/D, I/O	I/O	Buffered			17			26		ns

†All typical values are at  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

**PRELOAD PROCEDURES**

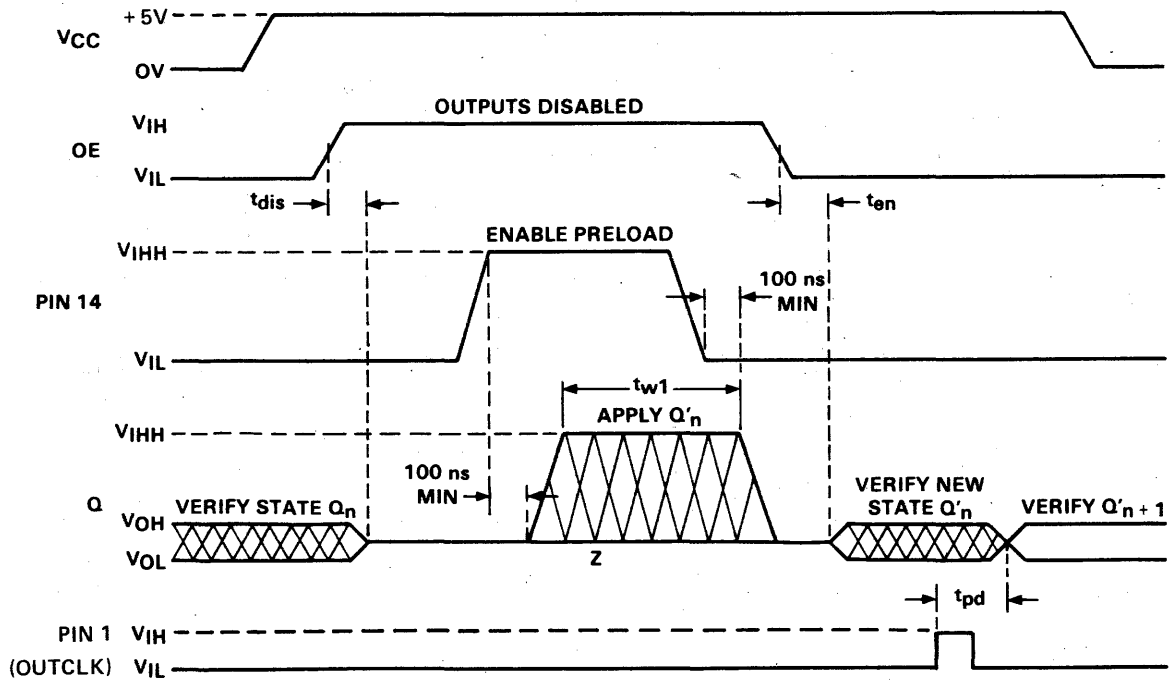


FIGURE 1—PRELOAD WAVEFORMS

**preload procedure for registered outputs**

- Step 1 Pin 13 to  $V_{IH}$ , Pin 1 to  $V_{IL}$ , and  $V_{CC}$  to 5 volts.
- Step 2 Pin 14 to  $V_{IHH}$ .
- Step 3 Apply an open circuit for a low and  $V_{IHH}$  for a high at the Q outputs.
- Step 4 Pin 14 to  $V_{IL}$ .
- Step 5 Remove the voltages applied to the outputs.
- Step 6 Pin 13 to  $V_{IL}$ .
- Step 7 Check the output states to verify preload.



**FIELD-PROGRAMMABLE LOGIC**

**TYPES SN54PLT19L8, SN54PLT19R4, SN54PLT19R6, SN54PLT19R8  
SN74PLT19L8, SN74PLT19R4, SN74PLT19R6, SN74PLT19R8  
LATCHED-INPUT FIXED-OR ARRAYS**

D2710, DECEMBER 1982—REVISED SEPTEMBER 1983

- Standard 24-Pin, 300-mil Packages
- Output Registers Automatically Clear During Power-Up
- Output Registers Have Preload Capability
- Data Input Registers Programmable to Buffers
- Choice of Operating Speeds
  - 1 Parts . . . 30 MHz Max, Standard power
  - 2 Parts . . . 20 MHz Max, Half power

DEVICE	I/D INPUTS	I INPUTS	3-STATE 0 OUTPUTS	REGISTERED Q OUTPUTS	I/O PORTS
'PLT19L8	11	2	2	0	6
'PLT19R8	11	0	0	8 (3-state buffers)	0
'PLT19R6	11	0	0	6 (3-state buffers)	2
'PLT19R4	11	0	0	4 (3-state buffers)	4

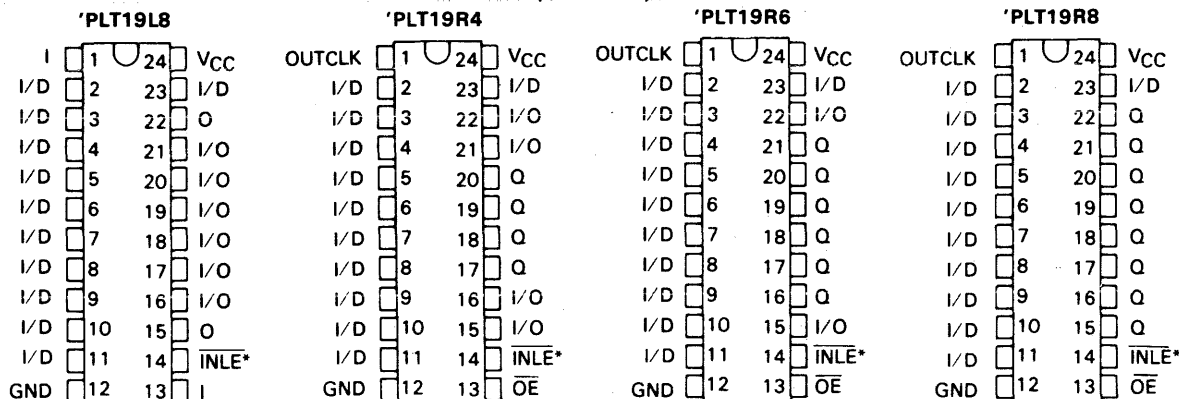
**description**

These fixed-OR arrays with eleven data inputs provide input registers that can be used as they are or be programmed into buffers. Some outputs of the 'PLT19R8, 'PLT19R6, and 'PLT19R4 have registers that can be loaded from the I/O pins by a preload procedure, while others are I/O ports and standard 3-state outputs. All the outputs are automatically set to a low level when power is applied. The -1 and -2 parts offer a choice of operating frequency, switching times, and power dissipation.

The SN54PLT19' is characterized for operation over the full military temperature range of -55°C to 125°C. The SN74PLT19' is characterized for operation from 0°C to 70°C.

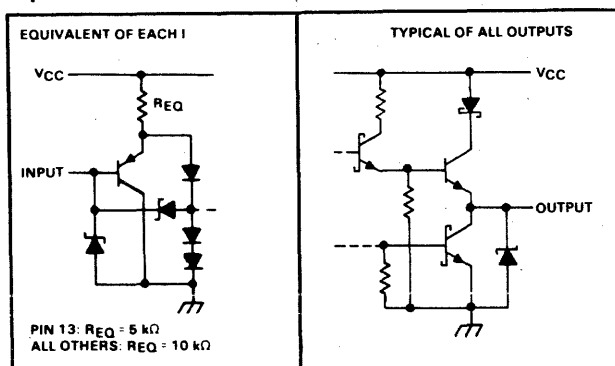
**pin assignments in operating mode (voltages at pins 1 and 13 less than V<sub>I(H)</sub>)**

SN54' ... JT PACKAGE  
SN74' ... NT PACKAGE  
(TOP VIEW)



\*Pin 14 is also used for the preload procedure on the last page of this data sheet.

**schematics of inputs and outputs**



**ADVANCE INFORMATION**

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This document contains information on a new product. Specifications are subject to change without notice.

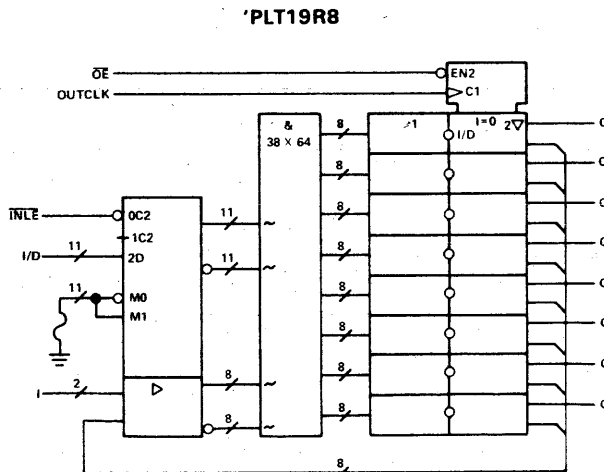
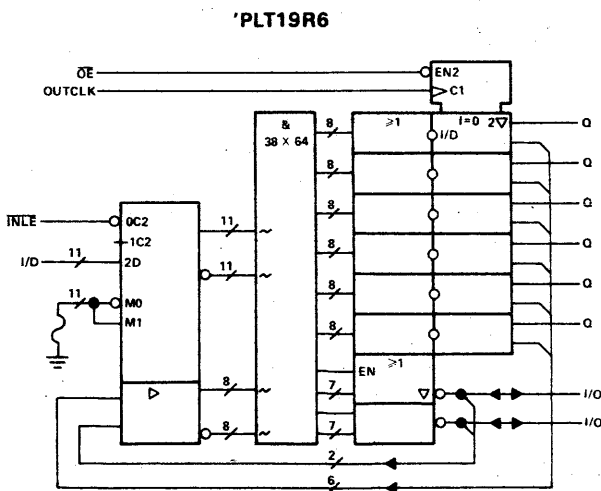
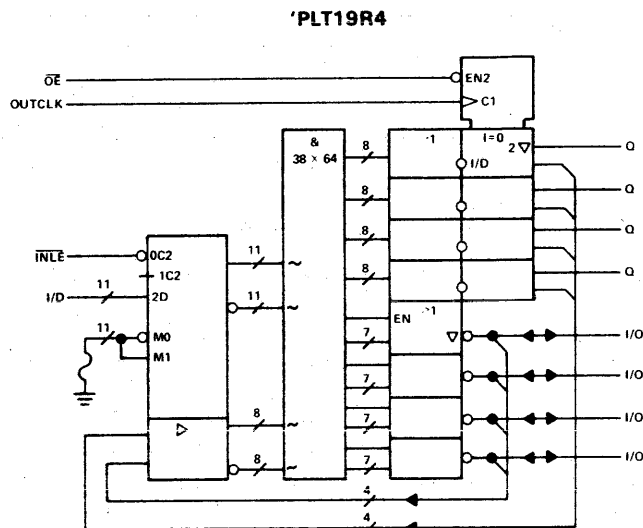
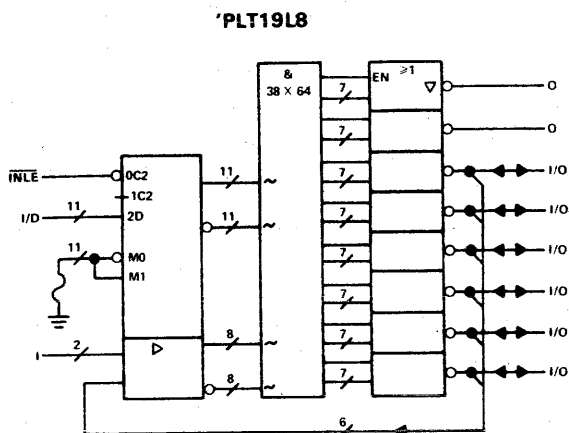
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# TYPES SN54PLT19L8, SN54PLT19R4, SN54PLT19R6, SN54PLT19R8 SN74PLT19L8, SN74PLT19R4, SN74PLT19R6, SN74PLT19R8 LATCHED-INPUT FIXED-OR ARRAYS

## functional block diagrams



### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC}$ (see Note 1) .....	7 V
Input voltage (see Note 1) .....	5.5 V
Voltage applied to a disabled output (see Note 1) .....	5.5 V
Voltage at any programming pin .....	12 V
Operating free-air temperature range: SN54PLT' .....	-55°C to 125°C
SN74PLT' .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C

NOTE 1: These ratings apply except for programming pins during a programming cycle or during a preload cycle.

See Application Note section for additional information on Texas Instruments Bipolar Programmable Logic Arrays.

**TYPES SN54PLT19L8, SN54PLT19R4, SN54PLT19R6, SN54PLT19R8,  
SN74PLT19L8, SN74PLT19R4, SN74PLT19R6, SN74PLT19R8  
LATCHED-INPUT FIXED-OR ARRAYS**

**recommended operating conditions**

		SN54PLT19'			SN74PLT19'			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
V <sub>CC</sub>	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
V <sub>IH</sub>	High-level input voltage	2		5.5	2		5.5	V
V <sub>IL</sub>	Low-level input voltage			0.8			0.8	V
V <sub>OH</sub>	High-level output voltage			5.5			5.5	V
I <sub>OH</sub>	High-level output current			-2			-3.2	mA
I <sub>OL</sub>	Low-level output current			12			24	mA
T <sub>A</sub>	Operating free-air temperature	-55		125	0		70	°C

**electrical characteristics over recommended free-air operating temperature range**

PARAMETER	TEST CONDITIONS†	SN54PLT19'			SN74PLT19'			UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>IK</sub>	V <sub>CC</sub> = MIN, I <sub>I</sub> = -18 mA			-1.5			-1.5	V
V <sub>OH</sub>	V <sub>CC</sub> = MIN, I <sub>OH</sub> = MAX	2.4	3.2		2.4	3.3		V
V <sub>OL</sub>	V <sub>CC</sub> = MIN, I <sub>OL</sub> = MAX		0.25	0.4		0.35	0.5	V
I <sub>OZH</sub>	Outputs			20			20	μA
	I/O ports	V <sub>CC</sub> = MAX, V <sub>O</sub> = 2.7 V		100			100	
I <sub>OZL</sub>	Outputs			-20			-20	μA
	I/O ports	V <sub>CC</sub> = MAX, V <sub>O</sub> = 0.4 V		-250			-250	
I <sub>I</sub>	OE Input			0.2			0.2	mA
	All others	V <sub>CC</sub> = MAX, V <sub>I</sub> = 5.5 V		0.1			0.1	
I <sub>IH</sub>	OE Input			40			40	μA
	All others	V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.7 V		20			0.1	
I <sub>IL</sub>	OE Input			-0.4			-0.4	mA
	All others	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V		-0.2			-0.2	
I <sub>O</sub> §	V <sub>CC</sub> = MAX, V <sub>O</sub> = 2.25 V	-30		-125	-30		-125	mA
I <sub>CC</sub>	-1 Parts		150	200		150	200	mA
	-2 Parts	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0 V, Outputs open	75	100		75	100	

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡All typical values are V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

§The output conditions have been chosen to produce a current that closely approximates one half the true short-circuit current, I<sub>OS</sub>.

**'PLT19R8, 'PLT19R6, 'PLT19R4 timing requirements**

		-1 PARTS		-2 PARTS		UNIT
		MIN	MAX	MIN	MAX	
f <sub>clock</sub>	Clock frequency	0	30	0	30	MHz
t <sub>w</sub>	Clock pulse duration, clock high or low	12		12		ns
t <sub>su</sub>	Setup time, D input before INLE†					ns
t <sub>su</sub>	Setup time, input or feedback before OUTCLK†	15		15		ns
t <sub>h</sub>	Hold time, input or feedback before OUTCLK†	0		0		ns

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**TYPES SN54PLT19L8, SN54PLT19R4, SN54PLT19R6, SN54PLT19R8,  
SN74PLT19L8, SN74PLT19R4, SN74PLT19R6, SN74PLT19R8  
LATCHED-INPUT FIXED-OR ARRAYS**

switching characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	FROM	TO	INPUT MODE	TEST CONDITIONS	- 1 PARTS			- 2 PARTS			UNIT
					MIN	TYP†	MAX	MIN	TYP†	MAX	
$f_{max}$			Either		30			20			MHz
$t_{pd}$	I, I/O	I/O, O	Either			16		25			ns
$t_{pd}$	OUTCLK ↑	Q	Either			12		20			ns
$t_{en}$	$\overline{OE} \downarrow$	Q	Either			8		15			ns
$t_{dis}$	$\overline{OE} \uparrow$	Q	Either			6		12			ns
$t_{pd}$	$\overline{INLE}$	I/O, O	Latched			16		25			ns
$t_{en}$	$\overline{INLE}$	I/O, O, Q	Latched			25		35			ns
$t_{dis}$	$\overline{INLE}$	I/O, O, Q	Latched			20		30			ns
$t_{pd}$	I/D	I/O, O	Buffered			20		30			ns
$t_{en}$	I/D, I/O	I/O	Buffered			22		32			ns
$t_{dis}$	I/D, I/O	I/O	Buffered			17		26			ns

†All typical values are  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$ .

**PRELOAD PROCEDURES**

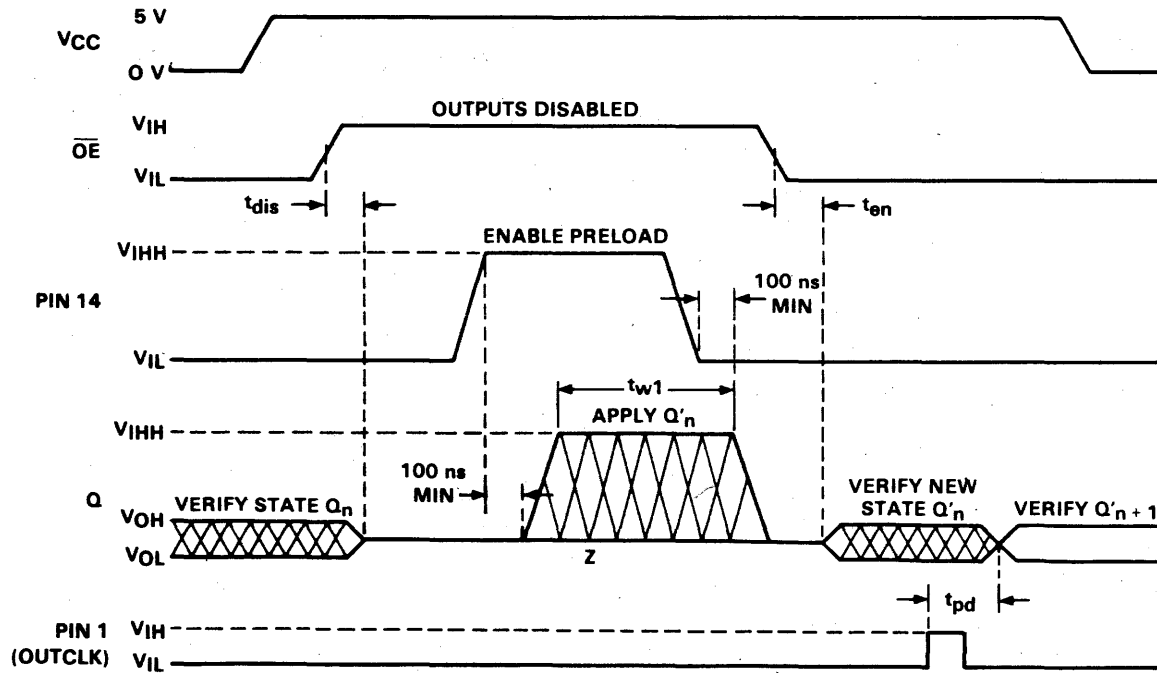


FIGURE 1—PRELOAD VOLTAGE WAVEFORMS

**preload procedure for registered outputs**

- Step 1 Pin 13 to  $V_{IH}$ , Pin 1 to  $V_{IL}$ , and  $V_{CC}$  to 5 volts.
- Step 2 Pin 14 to  $V_{IHH}$ .
- Step 3 Apply an open circuit for a low and  $V_{IHH}$  for a high at the Q outputs.
- Step 4 Pin 14 to  $V_{IL}$ .
- Step 5 Remove the voltages applied to the outputs.
- Step 6 Pin 13 to  $V_{IL}$ .
- Step 7 Check the output states to verify preload.

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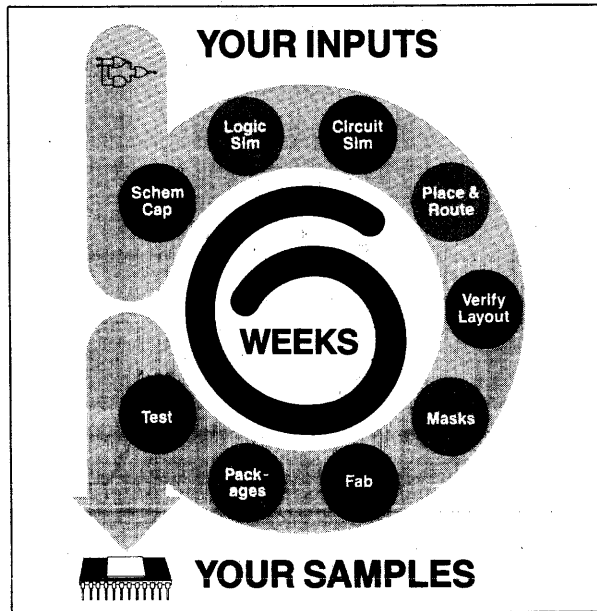
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- USER-DESIGNED VLSI** • Our complete and integrated set of CAE/CAD IC design tools provides you with everything you need to design your own ICs. They're easy-to-learn. And they'll help you get your design into silicon today, way ahead of your competition.
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### VTI'S SILICON FOUNDRY: WE CALL IT THE SILICON EXPRESS™

So where do you go once you have your custom IC design? How do you turn it into silicon? Before, you had to deal with fabrication services that had high volume requirements. But VTI has changed that for good.

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- QUICK TURNS** • Usually 14 days from mask entry to wafers out, 21 days to packaged prototypes.
- ADVANCED PROCESSES** • State-of-the-art CMOS, HMOS and NMOS in a variety of process options.
- TECHNOLOGY COMPATIBILITY** • Assured for custom ICs designed either with VTI's integrated CAE/CAD system or your own CAD tools.
- VTInet™** • The electronic foundry network that links you directly with us. Send your CIF, pattern generator or data base files through VTInet. Transmit your data base within hours through a system with built-in safeguards that enforce security and ensure our proprietary obligations to you.

Bring us your designs or your tooling. VTI's Silicon Express—just think of us as your silicon manufacturing arm. We do.

### AVAILABLE PROCESSES

Rule	NMOS	HMOS I	HMOS II	CMOS
TRANSISTOR LENGTH (microns)	5	3	2	3
GATE OXIDE (angstroms)	850	650	400	500
SELECTIVE ENHANCEMENT IMPLANT	no	yes	yes	—
CMOS WELLS	—	—	—	N-WELL TWIN-WELL
CONTACT SIZES (microns)	4X4	3X3	2X2	3X3
METAL PITCH	9.0	7.5	6.0	7.0
DUAL METAL	—	1Q84	1Q84	1Q84

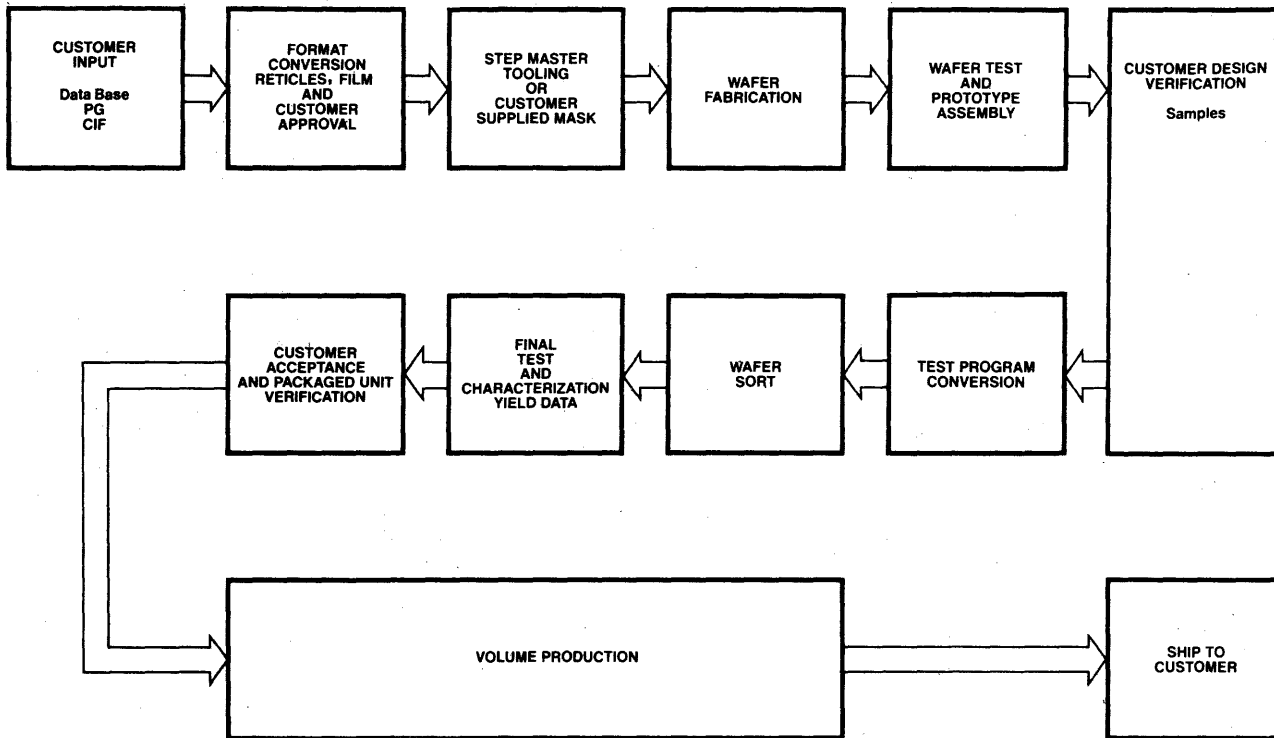
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# TRAILBLAZING THE VLSI FUTURE



## VTI CUSTOMER INTERFACE—THE SILICON EXPRESS



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- Chip planning and schematic capture
- Composition tools
- Cell Compiler libraries
- Logic and timing simulators

We've created a CAE/CAD VLSI design system that involves your specialized knowledge and unique requirements right from the inception of your custom chip. And without requiring MOS/VLSI design expertise. VTI's complete array of VLSI CAE/CAD tools are unparalleled in the industry. After all, we've based these tools on our extensive real-world experience in IC design and manufacture.

So you get the best of all worlds: a set of CAE/CAD tools designed by the premier group of VLSI CAE/CAD programmers in the semiconductor industry—refined and tested under real world conditions before they're released to you. With VTI, you get the high performance and efficiency, proprietary security and long-term economy associated with conventional full-custom design... but with the convenience, ease of use and fast turnaround you'd normally expected only from gate arrays.

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# TRAILBLAZING THE VLSI FUTURE

## VTI OFFERS A CHOICE OF DESIGN APPROACHES

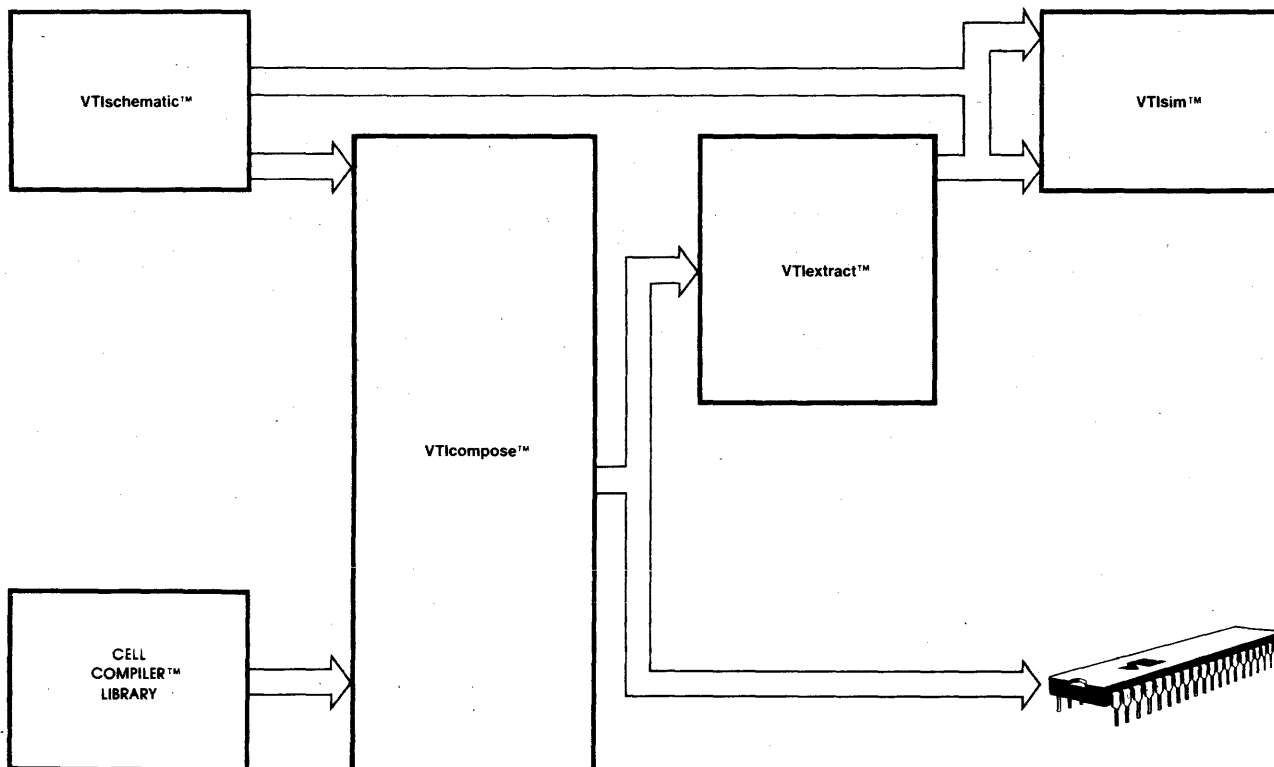
Design Approach	User	Benefits
LEVEL 1 LOGICOMP™	Logic and Systems Engineer	<ul style="list-style-type: none"> <li>▪ Easy to Learn</li> <li>▪ Very Rapid Design</li> </ul>
LEVEL 2 SYMCAD™	Systems and IC Designer	<ul style="list-style-type: none"> <li>▪ Unique Cell Design</li> <li>▪ Higher Performance</li> </ul>
LEVEL 3 GEOCAD™	Experienced IC Designer	<ul style="list-style-type: none"> <li>▪ Lowest Cost</li> <li>▪ Highest Performance</li> </ul>

### LOGICOMP: DESIGN TOOLS THAT SPEAK YOUR LANGUAGE

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- Enter a system schematic
- Construct a corresponding physical layout from predefined functional elements in VTI's Cell Compiler library
- Verify interconnection and function correctness of the physical layout, as specified by your schematic representation
- Automatically route all interconnects

### LOGICOMP LOGIC COMPILER™ SYSTEM



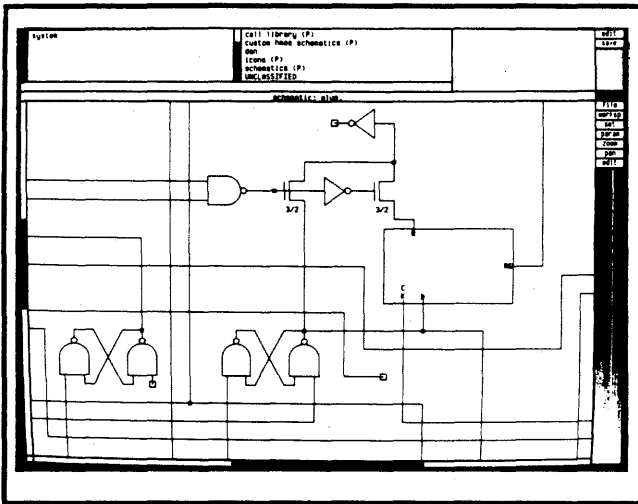
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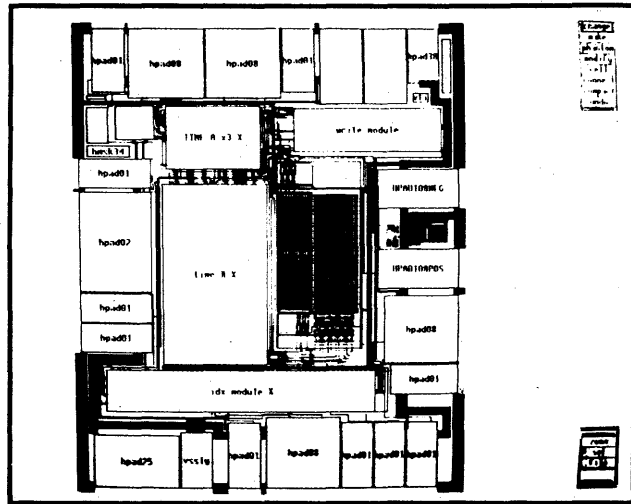




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### CELL COMPILERS: BUILDING BLOCKS FOR YOUR SYSTEM

VTI's unique Cell Compiler library—a concept that offers you ready-to-use functional elements to design your chip. Sounds like a standard cell library? Most people think so too, at first. Until they realize that VTI's Cell Compiler library offers the most advanced alterable set of cells available. Our Primary and Extended cell libraries compile physical layout databases for simple and complex IC building blocks. And we offer them in CMOS as well as HMOS. You specify the cell characteristics before compiling to physical layout. So you get a library with thousands of cell variations. Far more than the few hundred that conventional standard cell libraries offer.

With VTI's Cell Compilers, you specify your IC's unique requirements such as circuit speed, drive capability, data path width and functional repetition. What's more, with automatic cell layout generation, you don't need first-hand knowledge of MOS design rules or specific process details.

### CELL COMPILER FUNCTIONS

#### VTS810: Primary Library

- AND/NAND Gates
- Clock Generators
- Exclusive OR/NOR Gates
- Flip-Flops
- Input Buffers
- Inverting Buffers
- OR/NOR Gates
- Quasi-Static Latches
- 3-State Buffers
- TTL I/O Buffers

#### VTS820: Extended Library

- Adders/Subtractors
- ALUs
- Counters
- Data Demultiplexers
- Data Multiplexers
- Full Adders
- Pad-Ring Generators
- PLA Generators
- RAM Generators
- ROM Generators
- Shift Registers
- Up/Down Counters

### SYMCAD: BETTER CAD, LOWER COST

Your system design requires an IC made up of cells that simply don't exist. You could settle for a chip that's been glued together and works; but you'd like to design your own cells. You want a custom VLSI circuit that performs exactly the way you want it to, consumes the minimal power and takes only as much (or as little) room as necessary.

Obviously, you want access to a custom VLSI CAD system. Specifically, an interactive symbolic design package. And you'd prefer one without having to deal with awkward grids. Well, that system exists and is available to you. It's called SYMCAD: VTI's SYMBOLIC CAD system.

SYMCAD offers you:

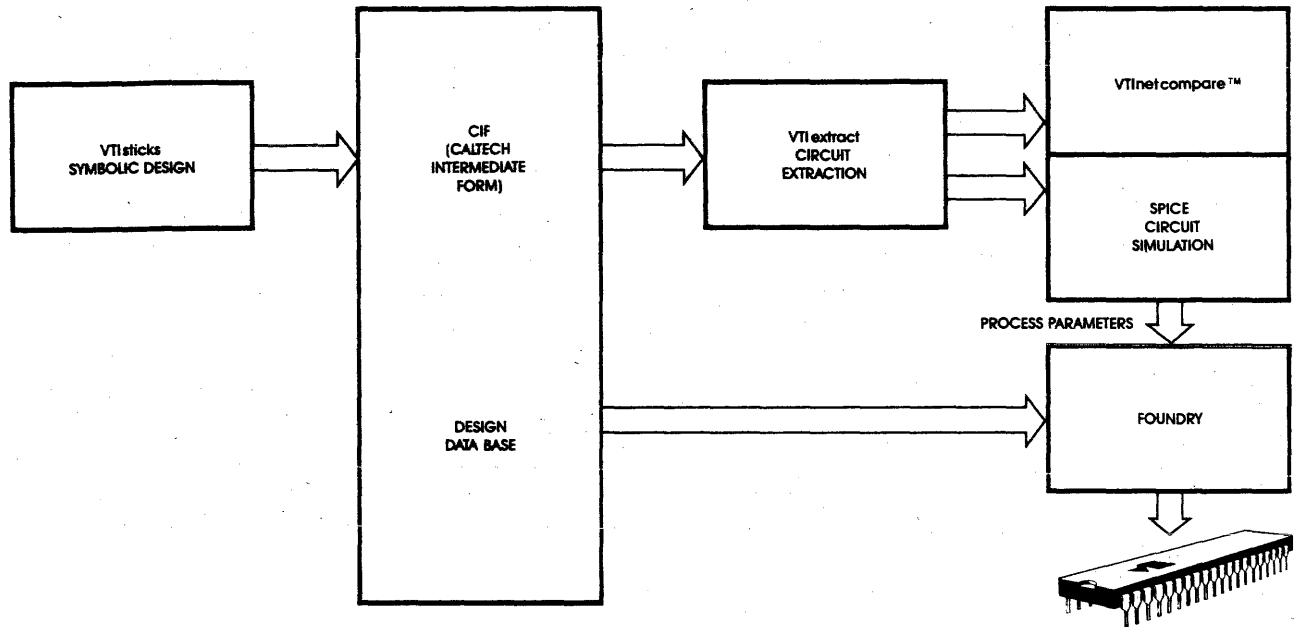
- VTisticks™ editor
- VTextract™—Circuit function and evaluation tool
- ERC—Electrical Rule Checker

A variety of features, integrally connected. The ideal way to design your system's components. SYMCAD: heralding your VLSI design future.

# TRAILBLAZING THE VLSI FUTURE



## SYMCAD SYMBOLIC COMPUTER-AIDED DESIGN SYSTEM



### VTisticks : THE KEY TO PRODUCTIVITY THROUGH SYMBOLIC DESIGN

You've been locked out of the IC design phase for your system. That is, up until today. Now, VTisticks opens the door that's kept you from designing the chip you need. VTisticks gives you total control over every facet of your system design. That's how it should have been. That's how it can be now. VTI's cell design tool automatically generates your IC layouts in compliance with user-specified design rules. Most importantly, the system is simple to use. You only have to specify the chip topology symbolically; then VTisticks converts your IC design to layout and compacts it to the minimal size. The benefit? Immediate custom MOS/LSI circuit design capability without detailed knowledge of MOS/LSI design techniques. And with assured results everytime.

#### VTisticks features:

- Automatic conversion from VTisticks to layout
- Automatic compaction to the minimum size allowed by design rules
- User-defined constraints on IC compaction
- Automatic recognition and placement of contacts
- Transistor size display and editing
- Symbolic naming

VTisticks automatically recognizes and places contacts between layers and transistors. And VTisticks' pop-up menus and mouse input provide you with an efficient way to enter your design. With VTisticks, you virtually eliminate keyboard entry in the layout phase of IC design.

With VTisticks, you'll create new cells that are nearly as dense as those laid out by hand. Even with those industry-unique features, VTisticks is easy-to-learn and easy-to-use. Your design will cost you less time, effort and money.



## **GEOCAD PHYSICAL LAYOUT EDITOR: GETTING THE SMALLEST CELLS POSSIBLE**

VTI's Physical Layout Editor, GEOCAD, is the premier interactive graphics tool for developing the most compact custom cells while maximizing performance. VTI allows you to design with arbitrary geometries. Once you've completed your design, the design rules can be verified with VTIdrc and the layout against the original logic design using VTExtract and VTIcompare.

Using GEOCAD is easy. You gain access through multiple windows whose sizes and positions you define. And then draw the graphics with a mouse.

The Layout Editor features:

- Any angle lines, any angle polygons
- Predefined and user-defined macros
- Edit-by-layer capability
- Cut, paste and edit commands, both within a window and between windows

## **VTI DESIGN CENTERS: PROVIDING DESIGN FACILITIES AND CONSULTING SERVICES**

We're putting all our design tools in your neighborhood—so you can reap the benefits of custom design. Use one of VTI's Design Centers as your own. We've staffed each of them with applications engineers, all of whom are ready to assist you in every way with chip design requirements and software support.

Our design centers are located at:

- 2235 Qume Drive  
San Jose, CA 95131  
408-943-0264
- 83 Cambridge Street  
Burlington, MA 01803  
617-229-6555
- 850 East Arapaho Road  
Richardson, TX 75081  
214-387-8350
- Via G. Jervis, 77  
10015 Ivera (To)  
Italy  
39-125-52-2010

You can come to our training classes. In less than two weeks, our VLSI Design Training Program will teach you the complexities and the simplicity of VTI's CAE/CAD design tools. In fact, by the end of your first week in Level 1, you'll be ready to design your own custom IC with VTI's unique Cell Compiler library. In Level 2, you'll learn the latest in MOS/VLSI design principles and the use of the VTIsticks graphics editor so you can design your own custom ICs. Our Level 3 course teaches you how to design your own Cell Compiler library as well as advanced layout editing techniques and advanced packaging and test considerations.

## **VLSI TRAINING, TOOLS AND SERVICES FROM VTI. INTENSIVE. FAST. COMPREHENSIVE.**

VLSI technology. An idea that's become a reality. With VTI, you'll take a step-by-step path to macrocell-based VLSI design nurtured by the experience and innovative prowess of a leading full-service VLSI semiconductor company. Whether you only need design training, our exciting design tools, or want the services of our designers or silicon foundry, you'll get the highest performance, the most efficient design combined with the latest knowledge in the semiconductor state-of-the-art.

After all, that's how VLSI technology is defined.

# ABBREVIATIONS OF COMPANY NAMES

<b>Action Ins</b>	Action Instruments	<b>GI</b>	General Instrument	<b>OAE</b>	Oliver Advanced Engineering
<b>AD</b>	Analog Devices	<b>GTE Micro</b>	GTE Microcircuits	<b>Octagon</b>	Octagon Systems Corp.
<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
<b>Advent</b>	Advent Products, Inc.	<b>Harris</b>	Harris Semiconductor	<b>Ohio Sci</b>	Ohio Scientific
<b>Alphatron</b>	Alphatron	<b>Heurikon</b>	Heurikon Corp.	<b>OKI</b>	OKI Semiconductor
<b>AMA</b>	American Automation	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hitachi</b>	Hitachi America, Ltd.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Holt</b>	Holt Inc.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>HP</b>	Hewlett-Packard		
<b>Analogic</b>	Analogic	<b>Hughes</b>	Hughes Aircraft, Solid State Products	<b>Panasonic</b>	Panasonic
<b>Analog Sys</b>	Analog Systems	<b>Hybrid Sys</b>	Hybrid Systems	<b>Pico Design</b>	Pico Design
<b>APC</b>	Applied Micro Circuits	<b>HyComp</b>	HyComp	<b>Polycore</b>	Polycore Electronics
<b>Apex</b>	Apex Microtechnology			<b>Plessey</b>	Plessey Semiconductors
<b>APM</b>	Applied Microsystems Corp.	<b>ICC</b>	International Cybernetics	<b>PMI</b>	Precision Monolithics, Inc.
<b>Appl Sys</b>	Applied Systems Corp.	<b>IDT</b>	Integrated Device Technology	<b>PragDes</b>	Pragmatic Design Inc.
<b>APT</b>	Applied Microtechnology	<b>IMI</b>	International Microcircuits, Inc.	<b>Pro-Log</b>	Pro-Log Corp.
<b>Aptek</b>	Aptek Microsystems	<b>IMP</b>	International Microelectronic Products		
<b>Array Tech</b>	Array Technology	<b>IMS</b>	Industrial MicroSystems Inc.	<b>Quay</b>	Quay Corp.
<b>AWI</b>	AWI Electronics	<b>Infosphere</b>	Infosphere		
		<b>Inmos</b>	Inmos	<b>Raytheon</b>	Raytheon Semiconductor
<b>Barvon</b>	Barvon Research	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>RCA</b>	RCA Solid State Division
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>IntCirSys</b>	Integrated Circuit Systems	<b>RCI Data</b>	RCI Data
<b>Burr-Brown</b>	Burr-Brown	<b>IntCompSys</b>	Integrated Computer Systems	<b>RELMS</b>	Relational Memory Systems
		<b>Int Tech</b>	Integrated Technology Corp.	<b>Reticon</b>	Reticon
<b>CAE</b>	Computer Aided Engineering	<b>Intech</b>	Intech Microcircuits Div.	<b>RIFA</b>	RIFA
<b>Cal Devices</b>	California Devices	<b>Intel</b>	Intel	<b>Rockwell</b>	Rockwell, Microelectronic Devices
<b>Cermetek</b>	Cermetek	<b>Interdesign</b>	Interdesign	<b>RTC</b>	Riehl Time Corporation
<b>CGRS</b>	CGRS Microtech Inc.	<b>Intersil</b>	Intersil		
<b>Cherry</b>	Cherry Semiconductor	<b>Intronics</b>	Intronics	<b>Sanyo</b>	Sanyo
<b>CIC</b>	Custom Integrated Circuits	<b>ITT</b>	ITT Semiconductors	<b>SBE</b>	SBE, Inc.
<b>CirTech</b>	Circuit Technology			<b>SEEQ</b>	SEEQ Technology, Inc.
<b>Citel</b>	Citel	<b>Kinetic Sys</b>	Kinetic Systems	<b>SPI</b>	Semi Processes Inc.
<b>Comlinear</b>	Comlinear Corporation	<b>Kontron</b>	Kontron Electronics	<b>Siemens</b>	Siemens
<b>CMA</b>	Custom MOS Arrays			<b>Si-Fab</b>	Si-Fab
<b>Comark</b>	Comark Corp.	<b>Lambda</b>	Lambda Semiconductor	<b>Signetics</b>	Signetics
<b>Comdial</b>	Comdial Semiconductor	<b>Linear Tech</b>	Linear Technology	<b>SGS</b>	SGS Semiconductor
<b>Comp Auto</b>	Computer Automation	<b>LSI Comp</b>	LSI Computer Systems	<b>Sharp</b>	Sharp
<b>Compas</b>	Compas Microsystems	<b>LSI Logic</b>	LSI Logic Corporation	<b>Silicon G</b>	Silicon General
<b>Cont Logic</b>	Control Logic Inc.			<b>Siliconix</b>	Siliconix
<b>Control Sys</b>	Control Systems Microsystems Div.	<b>Master Logic</b>	Master Logic Corporation	<b>Silicon Sys</b>	Silicon Systems Inc.
<b>CreMicro</b>	Creative Micro Systems	<b>Matrix</b>	Matrix Corp.	<b>Siltronics</b>	Siltronics
<b>Cromemco</b>	Cromemco, Inc.	<b>Matrox</b>	Matrox Electronic Systems	<b>SMC</b>	Standard Microsystems Corp.
<b>Cubit</b>	Cubit Inc.	<b>Maxim</b>	Maxim Integrated Products	<b>S MOS</b>	S MOS Systems
<b>Curtis</b>	Curtis Electro Devices, Inc.	<b>MCC</b>	Micro-Computer Control	<b>Solarise</b>	Solarise Enterprises
<b>Cybernetic</b>	Cybernetic Micro Systems	<b>MCE</b>	MCE Electronics	<b>Solitron</b>	Solitron Devices
<b>Cybersys</b>	Cybersystems	<b>Micrel</b>	Micrel	<b>Sprague</b>	Sprague Electric Company
<b>Cybertek</b>	Cybertek Inc.	<b>Micro Innov</b>	Micro Innovators	<b>SSM</b>	Solid State Micro Technology for Music
		<b>Micropac</b>	Micropac Industries	<b>SSS</b>	Solid State Scientific
		<b>Micro Net</b>	Micro Networks	<b>Stag</b>	Stag Microsystems
<b>Data General</b>	Data General	<b>Micro Pwr</b>	Micro Power Systems	<b>STC</b>	Storage Technology Corp.
<b>Data I/O</b>	Data I/O	<b>Micro Sci</b>	Micro Sciences Corp.	<b>STD</b>	STD Microsystems
<b>Data Trans</b>	Data Translation	<b>Micro Tech</b>	Microcircuits Technology	<b>Struc Des</b>	Structured Design Inc.
<b>Datel</b>	Datel	<b>Micro-Link</b>	Micro-Link Corporation	<b>Stynetic</b>	Stynetic Systems
<b>Datricon</b>	Datricon Corporation	<b>Micron</b>	Micron Technology	<b>Sunrise</b>	Sunrise Electronics
<b>DDC</b>	Data Devices Corporation	<b>MillerTron</b>	MillerTronics	<b>Sunshine</b>	Sunshine Semiconductor
<b>DEC</b>	Digital Equipment Corporation	<b>Miller</b>	Miller Technology	<b>Supertex</b>	Supertex Inc.
<b>Die-Tech</b>	Die-Tech	<b>Mittel</b>	Mitel Semiconductor	<b>Symtek</b>	Symtek Corp.
<b>Digelec</b>	Digelec Corp.	<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Synertek</b>	Synertek
<b>Digitek</b>	Digitek, Inc.	<b>MMI</b>	Monolithic Memories, Inc.	<b>Sys Innov</b>	Systems Innovations
<b>Dionics</b>	Dionics Inc.				
<b>Dist Comp</b>	Distributed Computer Systems	<b>Mostek</b>	Monolithic Systems Corp.	<b>Tau Zero</b>	Tau Zero Inc.
<b>Divers Tech</b>	Diversified Technology	<b>Motorola</b>	Motorola	<b>Technitrol</b>	Technitrol
		<b>MRC</b>	Motorola Semiconductor	<b>Tektronix</b>	Tektronix
<b>E-HI</b>	E-H International, Inc.	<b>Murray</b>	MRC Systems	<b>Teledyne C</b>	Teledyne Crystalonics
<b>EDI</b>	Electronic Designs Inc.	<b>Monosil</b>	Murray Consulting	<b>Teledyne P</b>	Teledyne Philbrick
<b>Elind</b>	Elind Electronica Industriale			<b>Teledyne S</b>	Teledyne Semiconductor
<b>EMM-SESCO</b>	EEM-SESCO	<b>National</b>	National Semiconductor	<b>Telefunken</b>	Telefunken
<b>Emulogic</b>	Emulogic Inc.	<b>NCM</b>	NCM Corp.	<b>Telmos</b>	Telmos
<b>ETI Micro</b>	ETI Micro	<b>NCR</b>	NCR Corp., Microelectronics Division	<b>Teltone</b>	Teltone Corporation
<b>Exar</b>	Exar Integrated Systems	<b>NEC</b>	NEC Electronics	<b>TI</b>	Texas Instruments
<b>Exel</b>	Exel Microelectronics	<b>Nitron</b>	Nitron	<b>Third Domain</b>	Third Domain
				<b>Thomson-CSF</b>	Thomson-CSF Components Corp.
<b>Fairchild</b>	Fairchild			<b>Toshiba</b>	Toshiba America
<b>Ferranti</b>	Ferranti Electric			<b>Trans-Data</b>	Trans-Data
<b>Force</b>	Force Computers			<b>TRW</b>	TRW LSI Products
<b>Fujitsu A</b>	Fujitsu America				
<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.			<b>Unitrode</b>	Unitrode
				<b>Universal</b>	Universal Semiconductor, Inc.
				<b>Varix</b>	Varix Corp.
				<b>VLSI Design</b>	VLSI Design Associates
				<b>VTI</b>	VLSI Technology, Inc.
				<b>Votrax</b>	Votrax
				<b>Weitek</b>	Weitek Corporation
				<b>Western</b>	Western Digital
				<b>Wintek</b>	Wintek Corp.
				<b>Xicor</b>	Xicor, Inc.
				<b>Ycom</b>	Ycom
				<b>Zendex</b>	Zendex Corp.
				<b>Zilog</b>	Zilog
				<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrex</b>	Zytrex Corp.



## The Arrow Lines

### Manufacturers alphabetically

#### A

Alco  
Alpha  
AMD  
Amphenol  
Analog Devices  
Applied Digital Data Systems  
Arrow•ACS/2200  
Augat  
AVX

#### B

Beckman  
Belden  
Bussmann

#### C

Cambion  
Cannon ITT  
Cornell Dubilier  
Centralab  
Chicago Miniature  
C & K  
Clarostat  
CDC

Drives  
Media  
Corning  
Crydom  
CTS

#### D

Dataproducts  
Data Technology Corporation  
Data Type  
Dialight  
Dilog

#### E

Elec-trol

#### F

Fairchild

#### G

General Electric  
Capacitors  
Relays  
Semiconductors  
General Instrument  
Microelectronics  
Optoelectronics  
Semiconductors

#### H

Hi-G

#### I

IBM  
Inmos  
Integral Data Systems  
Intel  
Semiconductors  
Systems  
International Rectifier  
Intersil

#### L

Littelfuse  
Linear Technology Corporation

#### M

Magnecraft  
Mallory  
Mepco/Electra  
MicroPro  
Microsemiconductor  
MMI  
Mostek  
Murata/Erie

#### N

NEC  
Printers

#### O

OMTI

#### P

Prentice  
Power One

#### Q

Quantum

#### R

Raytheon  
RCA  
Rotron

#### S

Seagate Technology  
Seaelectro  
Selanar  
Siemens  
Signetics  
Spectra-Strip  
Spectrol  
Sprague  
Standard Power  
Struthers Dunn

#### T

TeleVideo  
Systems  
Terminals  
Texas Instruments, Inc.  
Connectors  
Semiconductors  
Systems  
Terminals & Printers  
Thermalloy  
Thomas & Betts  
TRW  
Capacitors  
Cinch  
IRC  
LSI  
Motors  
Optron  
Semiconductors  
UTC  
Tusonix

#### U

Unitrode

#### V

Verbatim



#### NORTHWEST

Denver, Colorado (303) 696-1111  
Portland, Oregon (503) 684-1690  
Sacramento, California (916) 925-7456  
Salt Lake City, Utah (801) 539-1135  
San Francisco, California (408) 745-6600  
Seattle, Washington (206) 643-4800

#### SOUTHWEST

Albuquerque, New Mexico (505) 243-4566  
Los Angeles, California (213) 701-7500  
Orange County, California (714) 838-5422  
Phoenix, Arizona (602) 968-4800  
San Diego, California (619) 565-4800

#### NORTH CENTRAL

Cedar Rapids, Iowa (319) 395-7230  
Chicago, Illinois (312) 397-3440  
Madison, Wisconsin (608) 273-4977  
Milwaukee, Wisconsin (414) 764-6600

Minneapolis, Minnesota (612) 830-1800  
St. Louis, Missouri (314) 567-6888

#### SOUTH CENTRAL

Austin, Texas (512) 835-4180  
Dallas, Texas (214) 386-7500  
Houston, Texas (713) 530-4700  
Tulsa, Oklahoma (918) 665-7700

#### MIDWEST

Cleveland, Ohio (216) 248-3990  
Dayton, Ohio (513) 435-5563  
Detroit, Michigan (313) 971-8220  
Grand Rapids, Michigan (616) 243-0912  
Indianapolis, Indiana (317) 243-9353  
Pittsburgh, Pennsylvania (412) 856-7000

#### NORTHEAST

Boston, Massachusetts (617) 933-8130  
Manchester, New Hampshire (603) 668-6968  
Hauppauge, New York (516) 231-1000

Rochester, New York (716) 275-0300  
Syracuse, New York (315) 652-1000  
Wallingford, Connecticut (203) 265-7741

#### MID-ATLANTIC

Baltimore, Maryland (301) 247-5200  
Fairfield, New Jersey (201) 575-5300  
Marlton, New Jersey (609) 596-8000  
Richmond, Virginia (804) 282-0413

#### SOUTHEAST

Atlanta, Georgia (404) 449-8252  
Fort Lauderdale, Florida (305) 776-7790  
Huntsville, Alabama (205) 882-2730  
Melbourne, Florida (305) 725-1480  
Raleigh, North Carolina (919) 876-3132  
Winston Salem, North Carolina (919) 725-8711

Arrow International TWX: (510) 224-6021

**Arrow...blue ribbon service**

# **MANUFACTURERS AND DISTRIBUTORS/ REPRESENTATIVES DIRECTORY**

The Manufacturers and Distributors/Representatives Directory is a comprehensive telephone directory of the entire IC industry. It places at your fingertips the names and phone numbers you need to know. It includes manufacturers, international sales offices, representatives and distributors; for manufacturers with data pages, it also includes domestic sales offices, representatives and distributors. The manufacturers' listings tell you where and whom to call to obtain technical and ordering information. It enables you to ask for the right person or phone extension. It gives you the terminology understood by the individual company's switchboard operators, e.g., in some companies you should ask for sales, in others, customer service. This directory will reduce frustration and save you time.

# IC MASTER

## Action Instruments

Action Instruments Company, Inc.  
8601 Aero Drive  
San Diego, California 92123  
619-279-5726

## Advanced Micro Devices

Advanced Micro Devices, Inc.  
901 Thompson Place  
Sunnyvale, California 94088  
408-732-2400  
TELEX: 34-6306

### Specific product information:

Bipolar Microprocessors ..... Leon Torban  
Bipolar Memory ..... Dave Wood  
Logic (MSI) ..... Chris King  
Hi-Rel Logic and Interface ..... Ron Marfil  
Microcomputer Board Information ..... Advanced Micro  
Devices, 3340 Scott Blvd., Santa Clara, CA 95051, 408-  
988-7777, Telex: 171-142  
MOS Microprocessors and Support Circuits ..... Ed Huber  
MOS Serial Memory, Mask-programmable ROMs and E-  
PROMS ..... David Starr  
MOS-RAMs ..... Bill DeMatteis (DRAM), Dick Bruneau  
(SRAM)  
Computer Interface ..... Gary Conner

### Application engineering:

Communications ..... Russ Apfel  
Microprocessors ..... Bill Harmon  
Memory and General Purpose Logic ..... Kris Rallapalli

### Literature:

Literature Dept.

### Price and delivery:

Contact AMD Sales  
Office, Rep. or Dist.

### Follow-up an order:

Customer Service Dept.

### All other information:

Corporate Communications Dept.

## Sales Office & Representatives

**AZ** Phoenix  
AMD, 602-242-4400

**CA** El Segundo  
AMD, 213-640-3210

**CA** Newport Beach  
AMD, 714-752-6262

**CA** San Diego  
AMD, 619-560-7030

**CA** Santa Clara  
AMD, 408-727-1300

**CA** Santa Clara  
I((c2)) Inc., 408-988-3400

**CA** South America, Mtn. View  
Intectra, 415-967-8818

**CA** Woodland Hills  
AMD, 213-992-4155

**CO** Denver  
AMD, 303-691-5100

**CT** Cheshire  
Scientific Comps., 203-272-2963

**FL** Ft. Lauderdale  
AMD, 305-484-8600

**FL** Largo  
AMD, 813-535-9811

**GA** Atlanta  
AMD, 404-449-7920

**ID** Meridian  
Intermountain Technology, 208-888-5708

**IL** Elk Grove Village  
Oasis Sales, 312-640-1850

**IL** Itasca  
AMD, 312-773-4422

**IN** Indianapolis  
S.A.I. Marketing Corp., 317-241-9276

**IA** Cedar Rapids  
Lorenz Sales, 319-377-4666

**KS** Overland Park  
Kebco Inc., 913-541-8431

**KS** Wichita  
Kebco, Inc., 316-733-2117, 316-733-1301

**MD** Dorsey  
AMD, 301-796-9310

**MA** Burlington  
AMD, 617-273-3970

**MI** Brighton  
S.A.I. Mktg. Corp., 313-227-1786

**MN** Minnetonka  
AMD, 612-938-0001

**MO** Maryland Heights  
Kebco Inc., 314-576-4111

**NJ** Bellmawr  
T.A.I. Corp., 609-933-2600

**NJ** Edison  
AMD, 201-985-6800

**NH** Albuquerque  
Thorson Desert States, 505-293-8555

**NY** East Syracuse  
Nycorp, Inc., 315-437-8343

**NY** Liverpool  
AMD, 315-457-5400

**NY** Peapackego  
AMD, 914-471-8180

**NY** Woodbury  
AMD, 516-364-8020

**NC** Charlotte  
AMD, 704-525-1875

**OH** Centerville  
Dolfuss-Root & Co., 513-433-6776

**OH** Columbus  
AMD, 614-891-6455

**OH** Strongsville  
Dolfuss-Root & Co., 216-238-0300

**OR** Portland  
AMD, 503-245-0080

**PA** Bridgeville  
Dolfuss-Root & Co., 412-221-4420

**PA** Hersham  
AMD, 215-441-8210

**TX** Austin  
AMD, 512-346-7830

**TX** Dallas  
AMD, 214-934-9099

**UT** Salt Lake City  
R((c2)) Mktg., 801-298-2631

**WA** Bellevue  
AMD, 206-455-3600

**WI** Brookfield  
Oasis Sales, Inc., 414-782-6660

**Can** Kanata, Ottawa, Ontario  
Vitel Elctns., 613-836-1776

**Can** Mississauga, Ontario  
Vitel Electronics, 416-676-9720

**Can** St. Laurent, Quebec  
Vitel Elctns., 514-331-7393

**Can** Willowdale, Ontario  
AMD, 416-224-5193

**Int'l** Australia, Sydney  
R & D Elctns., TEL: (61) 439-5488

**Int'l** Australia, Victoria  
R & D Elctns., TEL: (03) 288-8232/62

**Int'l** Austria, Voessendorf-Wien  
Kontron Ges.m.b.H., TEL: 040-68 29 50

**Int'l** Belgium, Bruxelles  
AMD, TEL: (02) 771 9993

**Int'l** Belgium, Bruxelles  
MCAB, TEL: 02-375-29-01

**Int'l** Belgium, Dugres  
MCA Tronix, TEL: (041) 362780/362795

**Int'l** Denmark, Copenhagen  
Advanced Elctn. of Denmark ApS., TEL: (01)  
19 44 33

**Int'l** Finland, Korava  
Telercas Oy, TEL: 358-0-248055

**Int'l** France, Buc  
YREL, TEL: 03-956-81-42

**Int'l** France, Gentilly  
RTF, TEL: 01-664-11-01

**Int'l** France, Le Chesnay  
A2M, TEL: (03) 954.91.13

**Int'l** France, Nantes  
Sonepar Electronique Quest, TEL: 40-20-4-81

**Int'l** France, Rangis Cedex  
AMD, S.A., TEL: (01) 687-36-66

**Int'l** Germany, Burgwedel  
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**Analogic**

Analogic Corporation  
Audubon Road  
Wakefield, Massachusetts 01880  
617-246-0300  
TWX: 710-348-0425

**Analog Systems**

Analog Systems Division  
P.O. Box 35879  
Tucson, Arizona 85740  
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- Int'l** Spain, Madrid  
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- Int'l** Sweden, Vallingby  
Svenski Teleindustri, TEL: 380320
- Int'l** Switzerland, Ruschlikon  
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- Int'l** Taiwan, Taipei  
Alpha Precision Instrmn., TEL: (02) 752-7111

**Int'l** Yugoslavia, Ljubljana  
Elektrotehna N. Sol O. Ljubljana, TEL: 061/329745

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California Devices Inc.  
282 Kinney Drive  
San Jose, California 95112  
408-295-3700  
TELEX: 352120

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- Int'l** England, Salisbury, Wiltshire  
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- Int'l** France, Peissy  
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- Int'l** Herzlia B, Israel  
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Sunnyvale, California 94086  
408-734-8150  
TWX: 910-379-6931

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Micro-Marketing (Elctns.) Ltd., TEL: 06285-29222
- Int'l** England, Camberley Surrey  
Dialogue Distribution Ltd., TEL: 0276-682001
- Int'l** England, Maldenhead, Berkshire  
Micromark Elctns., TEL: (0628) 72631
- Int'l** France, Palaiseau  
Franelec, TEL: 6-930-2005
- Int'l** Holland, Rotterdam  
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- Int'l** Italy, Milano  
Cefra S. R. L., TEL: (02) 23.52.64-23.60.154
- Int'l** Sweden, Vallingby  
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CGRS Inc.  
P.O. Box 102  
Langhorne, Pennsylvania 19047  
215-757-0284

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Cherry Semiconductor Corporation  
2000 South County Trail  
East Greenwich, Rhode Island 02818  
401-885-3600  
TELEX: WUI-6817157  
**Specific product information:**  
Marketing & Sales Department ..... 401 885-3600

#### Sales Office & Representatives

- Int'l** England, Harpenden  
Cherry Electrical Prods. Ltd., TEL: (058) 27-63100
- Int'l** Hong Kong, Wanchoi  
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- Int'l** Japan, Tokyo  
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- Int'l** Taiwan, Taipei  
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160 Smith Street  
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516-293-8686

#### Sales Office & Representatives

- Int'l** Israel, Herzlia  
Electronic Microsystems, Ltd., TEL: 972-3-52-55895
- Int'l** United Kingdom, Surrey  
Circuit Technology Inc., TEL: 44-0276-23684

### Citel

Citel Inc.  
3060 Raymond St.  
Santa Clara, California 95050  
408-727-6562

### Comark

Comark Corporation  
93 West St., P.O. Box 474  
Medfield, MA 02052  
617-359-8161

### Comlinear

Comlinear Corporation  
2468 East 9th Street  
Loveland, Colorado 80537  
303-669-9433  
**Application engineering:**  
Kurt Rental

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CAE Systems  
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Sunnyvale, California 94086  
408-745-1440

**Computer Automation**

Computer Automation  
NAKED MINI Division  
18651 Von Karman  
Irvine, California 92713  
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TWX: 910-595-1767

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**Literature:**  
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Ext. 581

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Finn Metric OY, TEL: 90-460-844
- Int'l** France, Boitespostale  
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Micronic Devices, TEL: 486170
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Showa Info. Sys., TEL: 403-7101
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- Int'l** Norway, Oslo  
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- Int'l** South Africa, Braamfontein  
Tran Sys. Pty Ltd., TEL: 39-6575
- Int'l** Sweden, Stockholm  
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Geveke Elektronik, TEL: 0211/773082

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Nine Tech Circle  
Natick, Massachusetts 01760  
617-655-1170

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Creative Micro Systems  
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Los Alamitos, California 90720  
213-493-2484

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Mountain View, California 94039  
415-964-7400  
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- CA** Palo Alto  
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- GA** Norcross  
Hall-Mark Elctrn., 404-447-8000
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- TX** Dallas  
Kierulff Elctrn., 214-343-2400
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Kierulff Elctrn., 713-530-7030
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Kierulff Elctrn., 801-973-6913
- WA** Tukwila  
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- WI** Oak Creek  
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Kierulff Elctrn., 414-784-8160

**Cubit**

Cubit Inc.  
190 So. Whisman Rd.  
Mountain View, CA 94041  
415-962-8237

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P.O. Box 4090  
Mountain View, California 94040  
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Systems Mktg., Inc., TEL: 03-254-2751
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- Int'l Sweden, Stockholm**  
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- Int'l West Germany, Munchen**  
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Huntsville, Alabama 35802  
205-883-4410

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Santa Clara, California 95050  
408-727-9965

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TELEX: 320290

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- intl** **India, Bangalore**  
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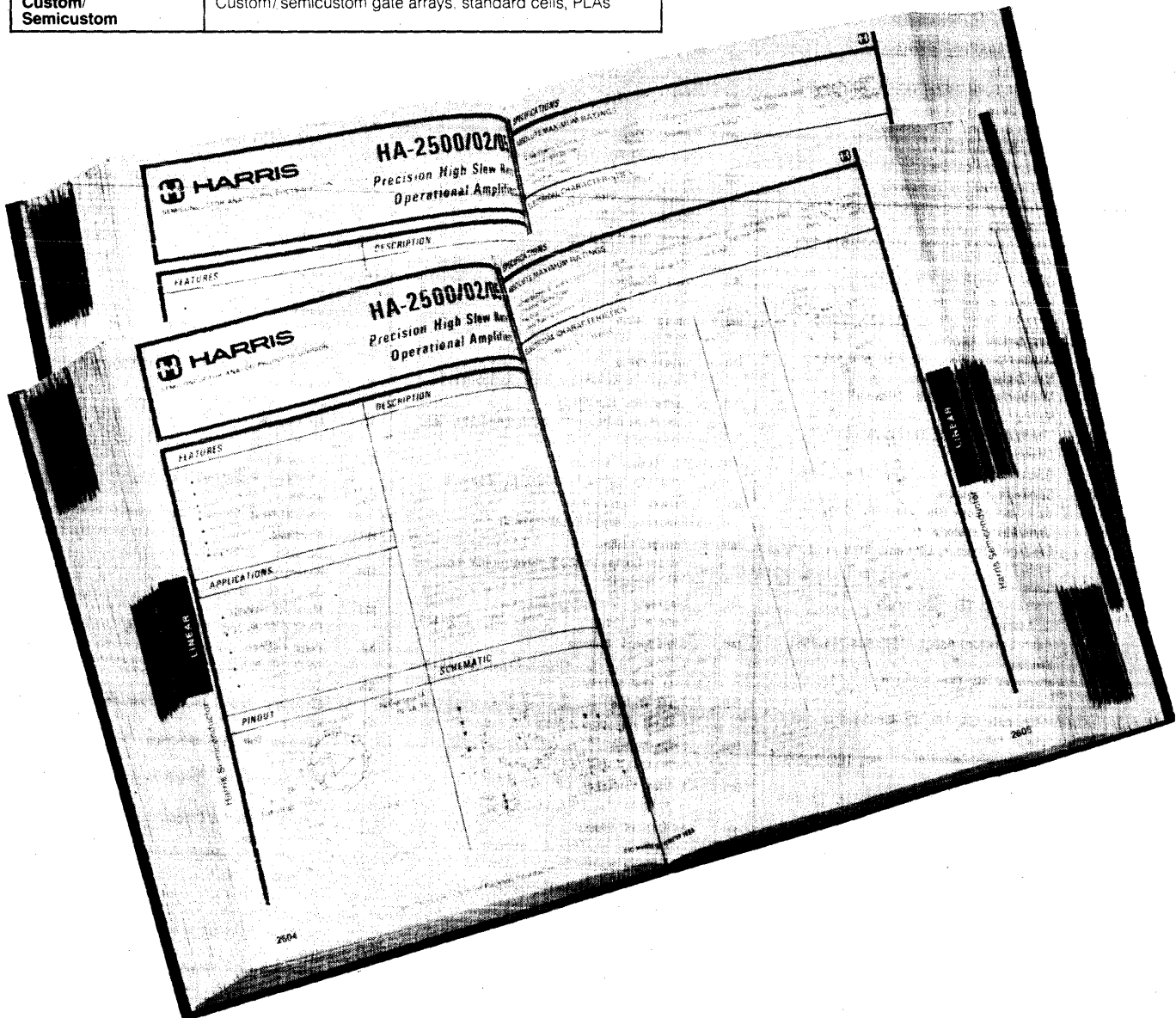


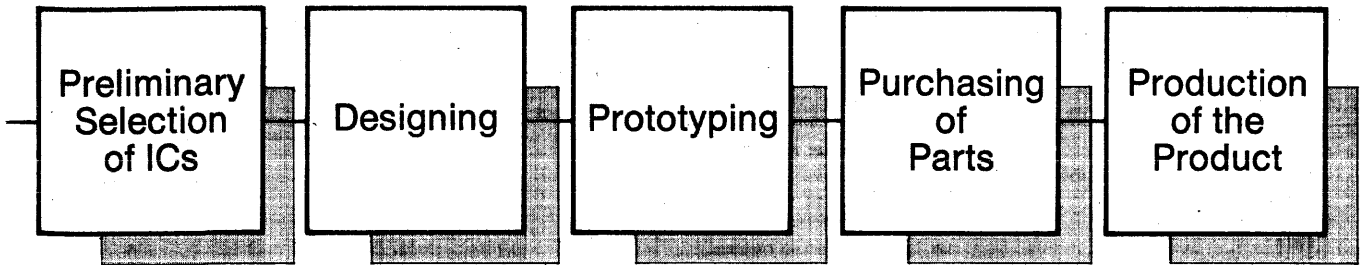
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## SIX KEY PHASES OF NEW SYSTEM DEVELOPMENT

Planning

# A good design





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Quality Comps., 214-387-4949

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Future Elctns., 514-694-7710

**Can** **Saskatoon, Saskatchewan**  
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Cam Gard Supply Ltd., 416-252-5031

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Cam Gard Supply Ltd., 604-291-1441

**Can** **Vancouver, British Columbia**  
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**Can** **Winnipeg, Manitoba**  
Cam Gard Supply Ltd., 204-786-8401

**Intl** **Finland, Helsinki**  
Yleiselektronikka/oy, TEL: 90-562 1122

**EXEL Microelectronics, Inc.**

EXEL Microelectronics, Inc.  
500 Valley Way  
Milpitas, California 95035  
408-942-0500

**Fairchild**

Fairchild Camera and Instrument Corp.  
Semiconductor Groups  
464 Ellis Street  
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415-962-5011  
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**Application engineering:**  
See Product Information

**Literature:**

Local field sales office or franchised distributor.

**Price and delivery:**

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**Place an order:**

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**Follow-up an order:**

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Fairchild Information Line, (415) 962-5011

**Sales Office & Representatives**

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**CA** **Santa Clara**  
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**FL** **Fort Lauderdale**  
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**IL** **Itasca**  
Field Sales Office, 312-773-3300-Disti, 312-773-3133-OEM

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**MN** **Edina**  
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- Can Nepean, Ontario**  
Field Sales Office, 416-665-9503
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- Intl Brazil, Sao Paulo**  
Fairchild Semiconductors, Ltda, TEL: 66-9092
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Fairchild Camera & Instrument S.A., TEL: 1-657-1303
- Intl Germany, Frankfurt Main**  
Fairchild Camera & Instrument GmbH, TEL: 49-611-6905613
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- Intl Italy, Roma**  
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Fairchild Semiconductor (Taiwan) Ltd., TEL: 573205
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Fairchild Camera & Instr. (UK) Ltd., TEL: 0707 51111

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Fairchild Camera & Instr., TEL: 07152 41026

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Ferranti Electric, Inc.  
87 Modular Avenue  
Commack, New York 11725  
516-543-0200  
TWX: 510-226-1490

**Specific product information:**  
Ken Kushman - IC Product Manager, Std IC's  
Anthony V. Walker - VLA Marketing Manager, Semi Custom I.C. Gate Arrays  
Joseph M. DeBlasio - VLA Marketing Engineer, Semi Custom I.C. Gate Arrays  
Helen A. Nelson - VLA Marketing Assistant, Semi Custom I.C. Gate Arrays  
John Day - VLA Systems Engineer, Semi Custom I.C. Gate Arrays  
Denzil Broadhurst - VLA Systems Engineer, Semi-Custom I.C. Gate Arrays  
John Carter - VLA Ststems Engineer, Semi Custom I.C. Gate Arrays

**Application engineering:**  
Ken Kushman

**Literature:**  
Jeanine Aiello

**Price and delivery:**  
Dina Gentile

**Follow-up an order:**  
Vivian Dennis

**All other information:**  
Ken Kushman

Sales Office & Representatives

- Intl United Kingdom, Chedderton-Didham**  
Ferranti, Ltd., TEL: (061) 624-0515
- Intl West Germany, Muenchen**  
Ferranti GmbH, TEL: (089) 293-871

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Fujitsu America  
910 Sherwood, Suite 23  
Lake Bluff, Illinois 60044  
312-295-2610

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- IL Elk Grove Village**  
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JMR Sales Inc., 516-689-8871
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Pro-Mark, 412-832-3088
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Whisler Sales Co., 206-643-1463

Fujitsu Microelectronics

Fujitsu Microelectronics, Inc.  
2985 Kifer Road  
Santa Clara, California 95051  
408-727-1700  
TWX: 910-338-0190

Specific product information:

Ed Alemany  
**Application engineering:**  
Bill Johnston  
**Literature:**  
Tim Henard  
**Price and delivery:**  
Respective sales office  
**Follow-up an order:**  
Respective sales office  
**All other information:**  
Ed Alemany (LD) 408-727-1700

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Milgray Connecticut, 203-795-0711

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NEP Electronics, 312-625-8400

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**MD Rockville**  
Milgray Washington, 301-468-6400

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**MA Burlington**  
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Carsten Electronics Ltd., 416-751-2371

**General Instrument**

General Instrument Corporation  
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600 W. John Street  
Hicksville, New York 11802  
516-733-3107  
TWX: 510-221-1866

**Sales Office & Representatives**

**Intl England, London**  
General Instrument Microelectronics Ltd.,  
TEL: 01-439-1895

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TEL: 01-439-1891
- intl France, Paris**  
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- intl Germany, Muenchen**  
General Instrument Deutschland GmbH, TEL:  
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General Instrument Hong Kong Ltd., TEL: (5)  
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- intl Japan, Tokyo**  
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- intl Taiwan, Taipei**  
General Instrument Microelectronics Taiwan,  
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- intl Austria, Wien**  
Elbatex GmbH, TEL: 0222/88 56 11
- intl Belgium, Bruxelles**  
Vekano, TEL: 7620505
- intl Denmark, Horlov**  
A/S Nordisk-Elektronik, TEL: 84.20.00
- intl England, London**  
Bee Arosales (Exporters), TEL: 01-636-6614/  
01-636 8211
- intl Finland, Espoo**  
Jorma Sarkkinen Ky., TEL: 046.10.88
- intl France, Boulogne**  
Gedis, TEL: 604.81.70

- intl France, Montrouge**  
P.E.P., TEL: 735.33.20
- intl Germany, Berlin**  
Roederstein-Baulemente Vertriebs GmbH, TEL:  
0 30/7 91 40 29
- intl Germany, Heilbronn**  
Elbatex GmbH, TEL: 0 71 31/8 90 01
- intl Germany, Muenchen**  
Electronic 2000 Vertriebs-GmbH, TEL: 0 89/  
43 40 61
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- intl Holland, Geldermalsen**  
Curijn Hasselaar, TEL: 03455-3150
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346.07
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Adrep Srl, TEL: (02) 408.41.01
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- intl Italy, Milan**  
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- intl United Kingdom, Slough**  
Crellon Electrns. Ltd., TEL: 06286 4434
- intl United Kingdom, Stevenage**  
Campbell Collins Ltd., TEL: 0438 69466
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Semiconductor Specialists Ltd., TEL: West  
Drayton 45522

General Micro Systems

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714-621-7532

GTE Microcircuits

GTE Microcircuits Division  
GTE Communications Products Corporation  
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602-968-4431  
TWX: 910-951-1383

Sales Office & Representatives

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Analog & Digital Bipolar/Digital CMOS Products Divisions/  
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P. O. Box 883  
Melbourne, Florida 32901  
305-724-7000  
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Field Sales Office

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**Specific product information:**  
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**Application engineering:**  
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- Clearwater, FL (813) 443-4514



**Distributors**

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**AL Huntsville**  
RM Electrn., 205-852-1550

**AZ Phoenix**  
Sterling Electronics, 602-258-4531

**AZ Scottsdale**  
Western Micro Technology, 602-948-4240

**AZ Tempe**  
Marshall Inds., 602-968-6181

**CA Anaheim**  
Time Electronics, 714-937-0911

**CA Canoga Park**  
Marshall Inds., 213-999-5001

**CA Chatsworth**  
Diplomat Electrn., 213-700-8700

**CA Chatsworth**  
Diplomat-Westates Electronics, 213-341-4411

**CA Chatsworth**  
IEC/Jaco, 213-998-2200

**CA Costa Mesa**  
Diplomat Electrn., 714-549-8401

**CA Cupertino**  
Western Micro Technology, 408-725-1660

# MANUFACTURERS & DISTRIBUTORS DIRECTORY

**Hitachi (Cont)**

**CA Irvine**  
Ryno Electronics, 714-261-2500

**CA Sacramento**  
IEC, 916-424-5297

**CA San Diego**  
CETEC Electronics, Inc., 619-278-5020

**CA San Diego**  
Diplomat Elctns, 619-292-5693

**CA San Diego**  
Marshall Inds., 619-578-9600

**CA San Diego**  
Ryno Electronics, 619-292-6022

**CA San Jose**  
CETEC Electronics, Inc., 408-263-7373

**CA Southgate**  
CETEC Electronics, Inc., 213-773-6521

**CA Sunnyvale**  
Diplomat Elctns, 408-734-1900

**CA Sunnyvale**  
Marshall Inds., 408-732-1100

**CA Sunnyvale**  
Time Elctns., 408-734-9888

**CA Torrance**  
Time Elctns., 213-320-0880

**CO Denver**  
IEC, 303-292-6121, 800-525-3367

**CO Denver**  
Marshall Inds., 303-427-1800

**CO Englewood**  
Diplomat Elctns., Inc., 303-740-8300

**CO Wheatridge**  
Active Component Technology, 303-422-9229

**CT Orange**  
Milgray Elctns., 203-795-0711

**CT Wallingford**  
Cronin Electronics, 203-265-3134

**FL Clearwater**  
Diplomat Elctns, 813-443-4514

**FL Fort Lauderdale**  
Time Elctns., 305-974-4800

**FL Orlando**  
Marshall Inds., 305-859-1620

**FL Palm Bay**  
Diplomat Elctns, 305-725-4520

**FL Tampa**  
Reptron Electronics, Inc., 813-855-4656

**FL Winter Park**  
Milgray Elctns., 305-647-5747

**GA Norcross**  
Diplomat Elctns, 404-449-4133

**GA Norcross**  
Marshall Industries, 404-934-5750

**IL Bensenville**  
Diplomat Elctns, 312-595-1000

**IL Hoffman Estates**  
IEC, 312-843-2040

**IL Lombard**  
RM Elctns., 312-932-5150

**IL Schaumburg**  
Marshall Inds., 312-490-0155

**IN Indianapolis**  
RM Elctns., 317-247-9701

**KS Overland Park**  
Milgray Elctns., 913-236-8800

**MD Baltimore**  
Resco/Baltimore, 301-823-0070

**MD Beltsville**  
Resco/Beltsville, 301-937-9100

**MD Columbia**  
Diplomat Elctns, 301-995-1226

**MD Columbia**  
Diplomat Elctns., Inc., 301-995-1226

**MD Gaithersburg**  
Marshall Industries, 301-840-9450

**MA Burlington**  
Milgray Electronics, 617-272-6800

**MA Needham**  
Cronin Electronics, 617-449-0500

**MA Norwood**  
Gerber Electronics Inc., 617-769-6000

**MA Westboro**  
Future Elctns., 617-366-2400

**MA Wilmington**  
R C Comps., Inc., 617-273-1860, 800-225-0818

**MA Woburn**  
Time Elctns., 617-935-8080

**MI Grand Rapids**  
RM Elctns., 616-531-9300

**MI Livonia**  
Marshall Industries, 313-525-5850

**MI Livonia**  
Reptron, 313-525-2700

**MN Bloomington**  
Semiconductor Specialists, Inc., 612-854-8841

**MN Plymouth**  
Marshall Inds., 612-559-2211

**MO Kansas City**  
Semiconductor Specialists, Inc., 816-452-3900

**MO Manchester**  
Time Elctns., 314-391-6444

**NJ Fairfield**  
Marshall Industries, 201-882-0320

**NJ Mt. Laurel**  
Marshall Industries, 609-234-9100

**NJ Perth Amboy**  
Sterling Elctns., 201-442-8000

**NJ Teaneck**  
Diplomat Elctns, 201-785-1830

**NM Albuquerque**  
Sterling Electronics, 505-884-1900

**NY Freeport**  
Milgray Elctns., 516-546-5600

**NY Hauppauge**  
Marshall Industries, 516-273-2424

**NY Hauppauge**  
Time Elctns., 516-273-0100

**NY Liverpool**  
Diplomat Elctns, 315-652-5000

**NY Melville**  
Diplomat Elctns, 516-454-6400

**NY Rochester**  
CAM/NY, 716-865-2080

**NC Raleigh**  
Resco/Raleigh, 919-781-5700

**OH Cleveland**  
CAM/RPC, 216-461-4700

**OH Cleveland**  
Milgray Elctns., 216-447-1520

**OH Dayton**  
Marshall Inds., 513-236-8088

**PA Broomall**  
Time Elctns., 215-359-1200

**PA Lancaster**  
CAM/RPC, 717-392-8141

**PA Philadelphia**  
Almo Electronics, 215-698-4000

**PA Pittsburgh**  
CAM/RPC, 412-782-3770

**TX Austin**  
Active Component Technology, 512-452-5254

**TX Austin**  
Marshall Industries, 512-458-5654

**TX Austin**  
Sterling Electronics, 512-836-1451

**TX Dallas**  
Active Comp. Technology, 214-980-1888

**TX Dallas**  
Marshall Inds., 214-233-5200

**TX Dallas**  
RM Elctns., 214-263-8361

**TX Dallas**  
Sterling Elctns., 214-243-1600

**TX Houston**  
Active Component Technology, 713-496-4000

**TX Houston**  
Marshall Inds., 713-777-0358

**TX Houston**  
Sterling Elctns., 713-627-9800

**UT No. Salt Lake City**  
IEC, 801-298-1869

**UT Salt Lake City**  
Active Component Technology, 801-487-8131

**UT Salt Lake City**  
Diplomat Elctns., 801-486-4134

**VA Richmond**  
Sterling Elctns., 804-266-2190

**WA Bellevue**  
IEC/JACO, 206-455-2727

**WA Bellevue**  
Marshall Inds., 206-747-9100

**WA Redmond**  
Western Micro Technology, 206-881-6737

**WI Milwaukee**  
Marsh Electronics, 414-475-6000

**WI New Berlin**  
RM Electronics, 414-784-4420

**Can Downsview, Ontario**  
Future Elctns., 416-663-5563

**Can Pointe Claire, Quebec**  
Future Elctns., 514-694-7710

**Can Vancouver, British Columbia**  
Future Elctns., 604-438-5545

**Holt**

Holt, Inc.  
8 Chrysler Street  
Irvine, California 92714  
714-859-8800  
TELEX: 182704

**Sales Office & Representatives**

**NY Great Neck**  
Cooper-Simon, 516-487-1142

**Hughes**

Hughes Aircraft Co.  
Solid State Products  
500 Superior Avenue  
Newport Beach, California 92663  
714-759-2942  
TWX: 910-596-1374

**Sales Office & Representatives**

**Int'l Kowdar Trivandram-695003, South India**  
Kryonix, TEL: 63805

**Distributors**

**Int'l Hong Kong, Kowloon**  
Tektron Elctns., Ltd., TEL: 3-856199

**Int'l Japan, Tokyo**  
Nihon Teksei Co., Ltd., TEL: (03) 461-5121

**Int'l Singapore, Beach Road**  
Tektron Elec. Ltd., TEL: 3-856199

**Hybrid Systems**

Hybrid Systems Corporation  
22 Linnell Circle  
Billerica, Massachusetts 01821  
617-667-8700  
TWX: 710-347-1575

Hybrid Systems (Cont)

Specific product information:

Henry Diamond, National Sales Manager ..... ext. 60  
 Art Berg, East Coast Regional Sales Manager ..... ext. 13  
 Jack Worthen, Northeast Regional Sales Manager ..... ext. 17  
 Charles J. Bevivino, West Coast Regional Sales Manager ..... ext. 18

Literature:

Cheryl Kane ..... ext. 11

Price and delivery:

Patricia G. Bradley ..... ext. 12

All other information:

Janet J. Ray ..... ext. 14

Product Marketing:

Director of Marketing, Barry Friedman ..... ext. 54

Sales Office & Representatives

- AL **Huntsville**  
Electronic Sales, Inc., 205-533-1735
- AZ **Phoenix**  
Relcom, 602-893-1209
- CA **Los Altos**  
Cain White & Co., 415-948-6533
- CA **San Marcos**  
Relcom, 714-741-9600
- CA **Thousand Oaks**  
Relcom, 805-529-4363
- CA **Walnut**  
Relcom, 714-598-9077
- CA **Woodland Hills**  
Relcom, 213-340-9143
- CO **Booulder**  
Star Sales, 303-440-4193
- CT **Enfield**  
MECO, Inc., 1-800-225-3409
- FL **Orlando**  
Delmac Sales, Inc., 305-898-4688
- FL **South Clearwater**  
Delmac Sales, Inc., 813-447-5192
- FL **Tamarac**  
Delmac Sales, Inc., 305-726-1330
- GA **Norcross**  
Electronic Sales, Inc., 404-448-6554
- IL **Arlington Heights**  
MJ Petersen & Assoc., Ltd., 312-577-8200
- KS **Olathe**  
Rush & West Assoc., 913-764-2700
- MD **Baltimore**  
Burgin Kreh Assocs., Inc., 301-265-8500
- MA **Reading**  
MECO, Inc., 617-944-6660
- MI **Wyoming**  
Berberich Assoc., Inc., 616-538-8879
- MN **Minneapolis**  
Northstar Comps., Inc., 612-553-1888
- MO **Ballwin**  
Rush & West Assoc., 314-394-7271
- NY **Commack**  
ERA, Inc., 516-543-0510
- NY **Syracuse**  
T-Squared Elctns., 315-463-8592
- NY **Victor**  
T-Squared Elctns., 716-924-9101
- NC **Raleigh**  
Burgin Kreh Assoc., Inc., 919-781-1100
- OH **Dayton**  
Berberich Assocs., Inc., 513-433-0342
- PA **Pittsburgh**  
Ron Moser Assoc., 412-931-0266
- TX **Austin**  
OM Sales Co., Inc., 512-452-1838
- TX **Dallas**  
OM Sales Co., Inc., 214-241-3876
- TX **Houston**  
OM Sales Co., Inc., 713-789-4426
- VA **Lynchburg**  
Burgin Kreh Assoc., Inc., 804-239-2626

- WI **Wauwatesa**  
Larson Assocs., Inc., 414-258-0529
- Can **Concord, Ontario**  
Kaytronics, 416-669-2262
- Can **Kanata, Ontario**  
Kaytronics, 613-592-6606
- Can **Surrey, British Columbia**  
Kaytronics, 604-581-7611
- Can **Ville St. Pierre, Quebec**  
Kaytronics, 514-367-0101
- Intl **Australia, Rozelle NSW**  
AMPEC Engrg. Co., TEL: 6128181166
- Intl **Austria, Wien**  
Transistor Vertriebsgesellschaft mbH, TEL: 829451
- Intl **Denmark, Copenhagen**  
Scansupply, TEL: 835090
- Intl **England, Camberley, Surrey**  
Hybrid Co. Comp. U.K. Ltd. Camberley, TEL: 0276/28128
- Intl **Finland, Helsinki**  
Carlo Casagrande Oy, TEL: 640711
- Intl **France, Rungis**  
Hybrid Sys. S.A.R.L., TEL: 33-1687-8336
- Intl **Germany, Darmstadt**  
Hybrid Sys. GmbH, TEL: 06151-291595
- Intl **Israel, Tel-Aviv**  
Vectronics Ltd., TEL: 23 44 24/ 22 84 72
- Intl **Italy, Milano**  
Tekelec/Airtronic, TEL: 73 80 641
- Intl **Japan, Tokyo**  
Tokyo Elec. Ltd. Panetron Div.
- Intl **Netherlands, Montfoort**  
Koning en Hartman, TEL: 210101
- Intl **Norway, Oslo**  
Elektronix A/S
- Intl **Sweden, Bromma**  
Ferner Elctns. AB
- Intl **Switzerland, Zurich**  
Industrade AG, TEL: 4131 363 2230

Hy Comp

Hy Comp Inc.  
 75 Union Avenue, Box 377  
 Sudbury, Massachusetts 01776  
 617-443-4631

Application engineering:

John J. Madden

Follow-up on order:

Customer Services:

Nancy Oelschlegel

All other information:

James Chacharone, V.P./Marketing Manager

ILC Data Device

ILC Data Device Corp.  
 105 Willbur Place  
 Bohemia, New York 11716  
 516-567-5600  
 TWX: 510-228-7324

Sales Office & Representatives

- Intl **Australia, Brookvale**  
Allied Capacitors Pty. Ltd., TEL: 77-9-011-61-2-938-2135
- Intl **England, Berkshire**  
DDC United Kingdom, Ltd., TEL: 44-488-82141
- Intl **England, London**  
ILC Data Device Corp., TEL: 44-1-3874689
- Intl **Finland, Helsinki**  
OY Suomen Bofors AB, TEL: 358-0-553-166
- Intl **France, Courbevoie**  
DDC Electronique, TEL: 33-1-333-5888

- Intl **Holland, Haarlem**  
Techmation Electronics B.V., TEL: 31-4189-2222
- Intl **India, Ludhiana**  
Inde Assoc., TEL: 37031
- Intl **Israel, Ramat Hasharon**  
R D T Elctns., TEL: 972-3-48.32.11
- Intl **Italy, Milan**  
Microelit Italia s.r.l., TEL: 39-2-469 0444
- Intl **Italy, Roma**  
Microelit Italia, s.r.l., TEL: 77-9-011-39-6-890892
- Intl **Japan, Tokyo**  
Minebea Co., Ltd., TEL: 77-9-011-81-3-763-6241
- Intl **Norway, Oslo**  
Henaco A/S, TEL: 77-9-011-47-2-162110
- Intl **Spain, Madrid**  
Anatronic, S.A., TEL: 34-1-2424455
- Intl **Sweden, Tyreso**  
Hybridteknik A/B, TEL: 46-8-7126720
- Intl **Switzerland, Baden-Daetwil**  
W. Stolz AG, TEL: 41-56-840151
- Intl **Taiwan, Taipei**  
Tai I Trading Co., Ltd., TEL: 886-2-713-2342
- Intl **West Germany, Muenchen**  
Fey Elektronik Bauelemente GmbH, TEL: 49-89-184041
- Intl **Yugoslavia, Titova**  
Elektrotehna, TEL: 61-329745

Industrial Micro-Systems

Industrial Micro-Systems Inc.  
 189 Hitchcock Road  
 Southington, Connecticut 06489  
 203-628-4844

Infosphere, Inc.

Infosphere, Inc.  
 4730 SW Macadam Ave.  
 Portland, Oregon 97201  
 503-226-3515

Distributors

- Intl **Japan**  
ASR-International, TEL: 602-949-8293
- Intl **New Zealand, Dunedin**  
K. J. Kirby & Assoc., TEL: (24) 34-580

Inmos

Inmos  
 P.O. Box 16000  
 Colorado Springs, Colorado 80935  
 303-630-4000  
 TWX: 910-920-4904

Specific product information:

Field Sales Offices

Application engineering:

Applications ..... 303-630-4493

Literature:

Field Sales Offices

Place an order:

Field Sales Offices/Distributors

All other information:

Marketing ..... 303-630-4361

Sales Office & Representatives

- AL **Huntsville**  
Macro-Marketing Assoc., 205-883-9630

# MANUFACTURERS & DISTRIBUTORS DIRECTORY

Inmos (Cont)					
AZ	<b>Phoenix</b> Compass Mktg. & Sales, 602-266-5400	WI	<b>Waukesha</b> Micro-Tex, 414-542-5352	NY	<b>Rochester</b> Arrow Elctns., 716-275-0300
CA	<b>Los Angeles</b> Ed Landa Co., 213-879-0770	Can	<b>Ottawa, Ontario</b> Har-Tech Electronics, 416-542-5352	NC	<b>Winston-Salem</b> Arrow Elctns., 919-725-8711
CA	<b>San Diego</b> Hadden Assoc., 619-565-9444	Can	<b>Pointe Claire, Quebec</b> Har-Tech Electronics, 514-697-6731	OH	<b>Centerville</b> Arrow Elctns., 513-435-5563
CA	<b>San Jose</b> Inmos Sales, 408-298-1786	Intl	<b>Italy, Milan</b> Cefra SRL, TEL: 392-235264	OH	<b>Solon</b> Arrow Elctns., 216-248-3990
CA	<b>Santa Clara</b> Criterion Sales, 408-988-6300	Intl	<b>Japan, Tokyo</b> Matsushita Electric, TEL: 03-435-4501	OK	<b>Tulsa</b> Arrow Elctns., 918-665-7700
CA	<b>Terrace</b> Inmos Sales, 213-530-7764	Intl	<b>United Kingdom, Bristol</b> Inmos Sales, TEL: 44-272-290-861	PA	<b>Horsham</b> Lionex Corp., 215-443-5150
CO	<b>Denver</b> Component Sales, 303-759-1666	Intl	<b>West Germany</b> Inmos Sales, TEL: 49-089-319-1028	PA	<b>Monroeville</b> Arrow Elctns., 412-856-7000
CT	<b>Yalesville</b> Kanan Assoc., 203-265-2404	<b>Distributors</b>		TX	<b>Addison</b> Quality Comps., 214-387-4949
FL	<b>Fern Park</b> E.I.R., Inc., 305-830-9600	AZ	<b>Tempe</b> Anthem Elctns., 602-244-0900	TX	<b>Austin</b> Arrow Elctns., 512-835-4180
GA	<b>Woodstock</b> Macro-Marketing Assoc., 205-883-9630	CA	<b>Chatsworth</b> Anthem Elctns., 213-700-1000	TX	<b>Austin</b> Quality Comps., 512-835-0220
IL	<b>Schaumburg</b> Micro-Tex, 312-885-1131	CA	<b>Chatsworth</b> Arrow Elctns., 213-701-7500	TX	<b>Dallas</b> Arrow Elctns., 214-386-7500
IN	<b>Indianapolis</b> Electro Reps, Inc., 317-842-7202	CA	<b>Newport Beach</b> Arrow Elctns., 714-851-8961	TX	<b>Stafford</b> Arrow Elctns., 713-530-4700
KS	<b>Kansas City</b> B.C. Electronic Sales, 913-342-1211	CA	<b>San Diego</b> Anthem Elctns., 619-453-4871	UT	<b>Salt Lake City</b> Arrow Elctns., 801-539-1135
KS	<b>Wichita</b> B.C. Electronic Sales, 316-942-9840	CA	<b>San Diego</b> Arrow Elctns., 619-565-4800	WA	<b>Bellevue</b> Arrow Elctns., 206-643-4800
MD	<b>Soverna Park</b> New Era Sales, 301-544-4100	CA	<b>San Jose</b> Anthem Elctns., 408-946-8000	WA	<b>Redmond</b> Anthem Elctns., 206-881-0850
MA	<b>Berlington</b> Inmos Sales, 617-273-5150	CA	<b>Sunnyvale</b> Arrow Elctns., 408-745-6600	WI	<b>Oak Creek</b> Arrow Elctns., 414-764-6600
MA	<b>Reading</b> Kanan Assoc., 617-944-8484	CA	<b>Testin</b> Anthem Elctns., 714-730-8000	Can	<b>Baxter Ctr., Ontario</b> Future Elctns., 613-820-8313
MI	<b>Grand Rapids</b> J. L. Montgomery Assoc., Inc., 616-774-9308	CO	<b>Denver</b> Arrow Elctns., 303-696-1111	Can	<b>Burnaby, B.C.</b> R-A-E, 604-291-8866
MI	<b>Southfield</b> J. L. Montgomery Assoc. Inc., 313-385-2616	CO	<b>Englewood</b> Anthem Elctns., 303-790-4500	Can	<b>Calgary, Alberta</b> Future Elctns., 403-259-6437
MN	<b>Edina</b> Mel Foster Tech Sales, 612-941-9790	CT	<b>Wallingford</b> Arrow Elctns., 203-265-7741	Can	<b>Downsview, Ontario</b> Future Elctns., 416-663-5563
MN	<b>Minneapolis</b> Inmos Sales, 612-831-5626	FL	<b>Ft. Lauderdale</b> Arrow Elctns., 305-776-7790	Can	<b>Edmonton, Alberta</b> R-A-E, 403-451-4001
MO	<b>St. Louis</b> B.C. Electronic Sales, 314-291-1101	FL	<b>Palm Bay</b> Arrow Elctns., 305-725-1480	Can	<b>Pointe Claire, Quebec</b> Future Elctns., 514-694-7710
NJ	<b>Teaneck</b> R.T. Reid Assoc., 201-692-0200	GA	<b>Norcross</b> Arrow Elctns., 404-449-8252	Can	<b>Vancouver, British Columbia</b> Future Elctns., 604-438-5545
NM	<b>Albuquerque</b> Compass Mktg. & Sales, 505-292-7377	IL	<b>Schaumburg</b> Arrow Elctns., 312-397-3440	Intl	<b>Belgium, Brussels</b> Diode, TEL: 322-4285105
NY	<b>E. Syracuse</b> L-Mar Assoc., 315-437-7779	IN	<b>Indianapolis</b> Arrow Elctns., 317-243-9353	Intl	<b>Denmark, Copenhagen</b> Advanced Elctn., TEL: 451-1-94433
NY	<b>Ithaca</b> L-Mar Assoc., 607-257-5712	MD	<b>Baltimore</b> Arrow Elctns., 301-247-5200	Intl	<b>Finland, Helsinki</b> Field Oy, TEL: 358-0 6922577
NY	<b>Rochester</b> L-Mar Assoc., 716-323-1000	MA	<b>Burlington</b> Lionex Corp., 617-272-9400	Intl	<b>Franca, Sarves</b> Tekelec Airtronic, TEL: 331-534-7535
NY	<b>W. Seneca</b> L-Mar Assoc., 716-675-1277	MA	<b>Webera</b> Arrow Elctns., 617-933-8130	Intl	<b>Israel</b> R.D.T. Elctns. Ltd., TEL: 972-3483211
NC	<b>Raleigh</b> Zucker Assoc., 919-782-8433	MI	<b>Ann Arbor</b> Arrow Elctns., 313-971-8220	Intl	<b>Italy, Milan</b> Cefra SRL, TEL: 392-235264
OH	<b>Beachwood</b> Thompson & Assoc., 216-831-6277	MN	<b>Edina</b> Arrow Elctns., 612-830-1800	Intl	<b>Netherlands, Badhoevedorp</b> Techmation Elctns., TEL: 31-4189-2222
OH	<b>Columbus</b> Thompson & Assoc., 614-272-2086	MO	<b>St. Louis</b> Arrow Elctns., 314-567-6888	Intl	<b>Norway, Oslo</b> Ola Tandberg Elektro, TEL: 472-197030
OH	<b>Dayton</b> Inmos Sales, 513-439-0988	NH	<b>Manchester</b> Arrow Elctns., 603-668-6968	Intl	<b>Spain, Madrid</b> Amitron, TEL: 341-24856863
OH	<b>Dayton</b> Thompson & Assoc., 513-435-7733	NJ	<b>Fairfield</b> Arrow Elctns., 201-575-5300	Intl	<b>Sweden, Stockholm</b> Gosta Backstrom, TEL: 468-541080
OR	<b>Berling</b> Westerberg & Assoc., 503-297-1719	NJ	<b>Fairfield</b> Lionex Corp., 201-227-7960	Intl	<b>Switzerland, Zurich</b> Omni Ray, TEL: 411-2520766
PA	<b>Huntingdon Valley</b> Knowles Assoc., 215-947-5641	NJ	<b>Moorestown</b> Arrow Elctns., 609-235-1900	Intl	<b>United Kingdom, High Wycombe, Bucks</b> Rapid Recall, TEL: 44-494-26271
TX	<b>Dallas</b> Nova Mktg., 214-750-6082	NM	<b>Albuquerque</b> Arrow Elctns., 505-243-4566	Intl	<b>United Kingdom, Sunbury on Thames, Middlesex</b> Hawke Elctns., TEL: 44-9327-85577
TX	<b>Houston</b> Nova Mktg., 713-988-6082	NY	<b>Hempstead</b> Arrow Elctns., 516-231-1000	Intl	<b>West Germany, Munich</b> EZ000 A.G., TEL: 49-89-434061
UT	<b>Salt Lake City</b> Component Sales, 801-466-8623	NY	<b>Hempstead</b> Lionex Corp., 516-273-1660	Intl	<b>West Germany, Quickborn</b> Astek Elctns., TEL: 49-4106-71084
WA	<b>Bellevue</b> Westerberg & Assoc., 206-453-8881	NY	<b>Liverpool</b> Arrow Elctns., 315-652-1000	Intl	<b>West Germany, Stuttgart</b> Ditronic, TEL: 49-711-722079
		NY	<b>Melville</b> Arrow Elctns., 516-694-6800		

# IC MASTER

## Intech Inc.

Intech Inc.  
2270 Martin Avenue  
Santa Clara, California 95050  
408-988-4390  
TWX: 910-338-2213

## Integrated Circuit Engineering

Integrated Circuit Engineering Corporation  
15022 North 75th Street  
Scottsdale, Arizona 85260  
602-998-9780

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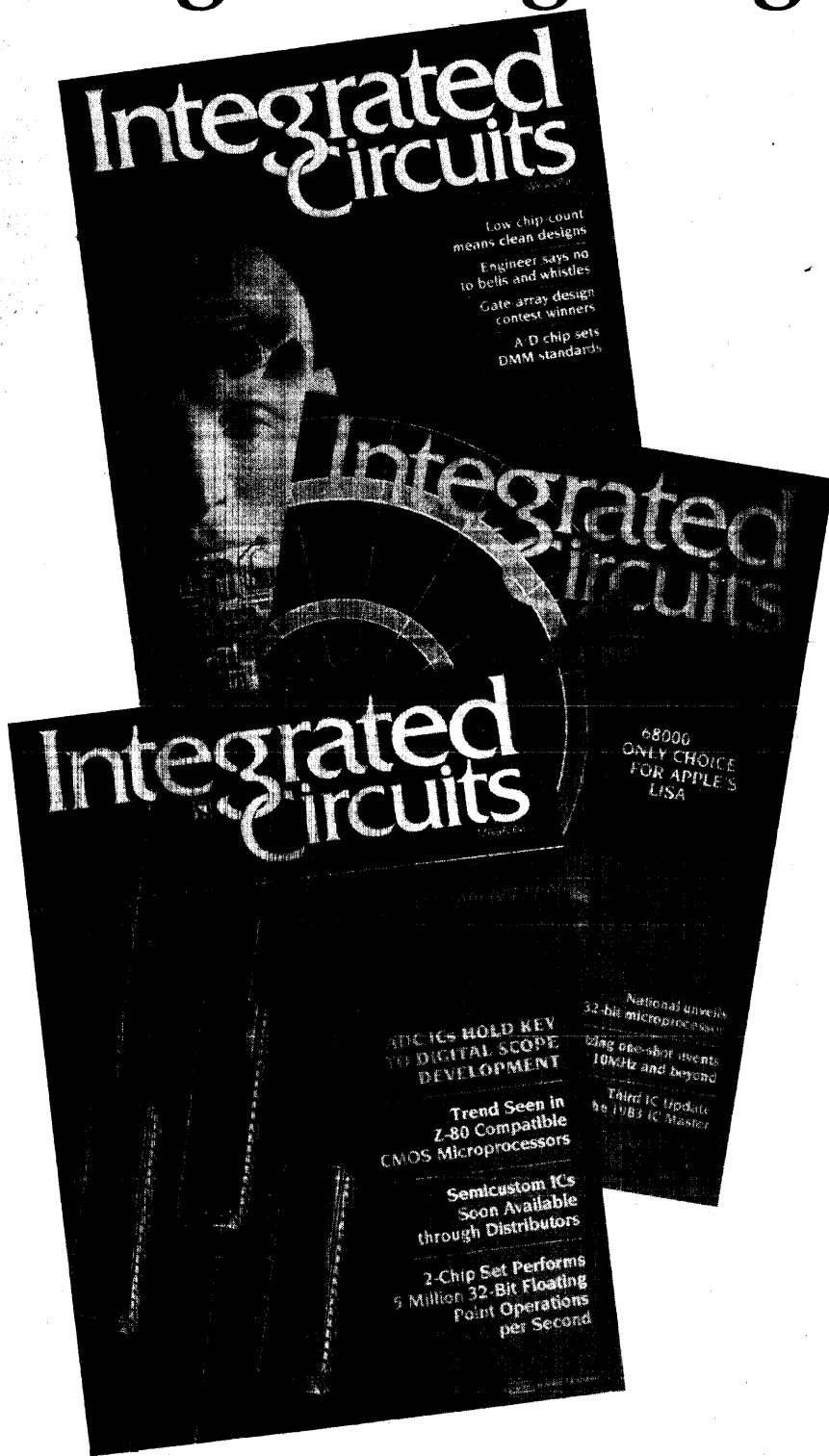
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MN	<b>Edina</b> Arrow Elctns., 612-830-1800	OR	<b>Lake Oswego</b> Hamilton/Avnet Elctns., 503-635-7848	Intl	<b>Chile, Santiago</b> Din, TEL: 227 564
MN	<b>Minnetonka</b> Hamilton/Avnet Elctns., 612-932-0666	PA	<b>Hersham</b> Pioneer/Delaware Valley, 215-674-4000	Intl	<b>China, Hong Kong</b> Schmidt & Co. Ltd., TEL: 75-8521-222
MN	<b>Minnetonka</b> Pioneer/Twin Cities, 612-935-5444	PA	<b>Monroeville</b> Arrow Elctns., 412-856-7000	Intl	<b>Colombia, Bogota</b> Intl. Computer Machines, TEL: 057-211-7282
MO	<b>Earth City</b> Hamilton/Avnet Elctns., 314-344-1200	PA	<b>Pittsburgh</b> Pioneer/Pittsburgh, 412-782-2300	Intl	<b>Denmark, Copenhagen</b> Scandinavian Semiconductor Supply A/S, TEL: (01) 83 50 90
MO	<b>St. Louis</b> Arrow Elctns., 314-567-6888	TX	<b>Austin</b> Hamilton/Avnet Elctns., 512-837-8911	Intl	<b>Denmark, Glostrup</b> MultiKomponent A/S, TEL: (02) 45 66 45
NH	<b>Manchester</b> Arrow Elctns., 603-668-6968	TX	<b>Austin</b> Pioneer/Austin, 512-835-4000	Intl	<b>Dublin, Ireland</b> Micro Marketing, TEL: (1) 85 62 88
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NJ	<b>Fairfield</b> Arrow, 201-575-5300	TX	<b>Dallas</b> Pioneer/Dallas, 214-386-7300	Intl	<b>France, Asnieres</b> Metrologie, TEL: (1) 791 44 44
NJ	<b>Fairfield</b> Hamilton, 201-573-3390	TX	<b>Houston</b> Hamilton/Avnet Elctns., 713-780-1771	Intl	<b>France, Bagnolet</b> Jermyn S.A., TEL: (1) 859 04 04
NJ	<b>Fairfield</b> Hamilton/Avnet Elctns., 201-575-3390	TX	<b>Houston</b> Pioneer/Houston, 713-988-5555	Intl	<b>France, Sevres</b> Tekelec Airtronic, TEL: (01) 534 75 35
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- intl Korea, Chung-Ku Seoul**  
Koram Digital, TEL: 782-8039
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Total Electronics, TEL: TELEX: 26297
- intl Les Ulis, France**  
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- intl United Kingdom, Berks**  
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- intl United Kingdom, Berkshire**  
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Jermyn Inds. (Mogul), TEL: (0732) 450144
- intl United Kingdom, Middlesax**  
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TWX: 910-598-4513

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Literature:  
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<b>Int'l</b>	<b>France, Buc</b> ERN, TEL: 956.00.11
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<b>Int'l</b>	<b>Japan, Tokyo</b> Internix, Inc., TEL: 03-369-1101
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<b>Int'l</b>	<b>United Kingdom, Birmingham</b> Steatite Microelectronics, TEL: 021 454 6961
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International Microsystems, Inc.  
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Auburn, California 95603  
916-885-7262

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**Application engineering:**

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**Literature:**

Marketing Service Department ..... 408-996-5403

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<b>CA</b>	<b>Sunnyvale</b> Intersil, Inc., 408-744-0618
<b>CA</b>	<b>Woodland Hills</b> Intersil, Inc., 213-887-7363
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<b>CT</b>	<b>Wallingford</b> COM-SALE, 203-269-7964
<b>FL</b>	<b>Fern Park</b> EIR, Inc., 305-830-9600
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<b>Int'l</b>	<b>Hong Kong</b> Intersil Datel, TEL: 3-7214286-7
<b>Int'l</b>	<b>West Germany, Muenchen</b> Intersil Datel GmbH, TEL: 89/530741

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Kontron Electronics, Inc.  
FutureData Division  
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Culver City, California 90230  
213-641-7200  
TWX: 910-328-7202

### Sales Office & Representatives

- Intl Austria, Wien**  
Kontron Elektronik GmbH & Co., TEL: 0043-222-670631-0
- Intl Belgium, Brussels**  
C.N. Rood B.V., TEL: 0032-2-7352135
- Intl Denmark, Hoersholm**  
E. Blichfeldt APS, TEL: 0045-2867900
- Intl Finland, Helsinki**  
OY Ammet AB, TEL: 0035-80-882044
- Intl France, Volzay-Villacoublay**  
Kontron Electronique, TEL: 00331-946722
- Intl Great Britain, St. Albans**  
Kontron Instruments Ltd., TEL: 0044-727-66222
- Intl Israel, Haifa**  
Tabor Electronics, TEL: 00972-4-644280

- Intl Italy, Cinisello Balsamo**  
Celdis, TEL: 0039-2-6120041
- Intl Italy, Milano**  
Telav International SRL, TEL: 0039-2-4455741
- Intl Netherlands, Rijswijk**  
C.N. Rood B.V., TEL: 0031-70-996360
- Intl Norway, Oslo**  
Bergman Instrumentering A/S, TEL: 0047-2-162210
- Intl Portugal, Lisboa Codex**  
Mattos Tavares Electronics Lda, TEL: 00351-19-616261
- Intl South Africa, Sandton**  
Biowave Electronics, SA, TEL: 0027-485808
- Intl Spain, Madrid**  
Belport Instrumentation SA, TEL: 00341-2628837
- Intl Spain, Madrid**  
Hispano Electronics SA, TEL: 0034-1-6194108
- Intl Sweden, Sandvberg**  
AB Transfer, TEL: 0045-8981620
- Intl Sweden, Upplands Vasby**  
Lagercrantz Elektronik AB, TEL: 0046-760-86120
- Intl Switzerland, Epalinges**  
Kontron Electronic AG, TEL: 0041-21331535
- Intl Switzerland, Zurich**  
Kontron instruments AG, TEL: 0041-1-4354111
- Intl West Germany, Echling bei Muechen**  
Kontron Messtechnik GmbH, TEL: 0049-89-31901-0

### Distributors

- Intl Australia, Melbourne**  
Parameters, TEL: 03-580-7444
- Intl Australia, Sydney**  
Parameters, TEL: 02-439-3288
- Intl Brazil, Sao Paulo, Rio, Porto Alegre**  
Cosele Instrumentos Electronicos, TEL: 255-1733
- Intl Hong Kong, Kwan Tong**  
Hong Kong Equipment Center, TEL: 3-413888
- Intl India, Bangalore**  
SRI RAM Associates
- Intl Korea, Seoul**  
Duksung Trading Co., TEL: 854-5047, 5831
- Intl New Zealand, Auckland**  
Professional Electronics, TEL: 493-029
- Intl Singapore, Singapore**  
Singapore Electronic & Engineering (Private) Ltd., TEL: 257-0211
- Intl Taiwan, Taipei**  
Gorman Co., Ltd., TEL: 02-5421924

## Lambda

Lambda Semiconductors  
121 International Drive  
Corpus Christi, Texas 78410  
512-289-0403 (In Texas); 1-800-255-9606  
TWX: 910-876-1404

### Sales Office & Representatives

- Intl England, High Wycombe, Bucks**  
Lambda Electronics, TEL: 44-494-36386
- Intl France, Les Ulis, Cedex**  
Lambda Electronique S.A., TEL: 33-6-012-1487
- Intl Germany, Achen**  
Lambda Netzgerate GmbH, TEL: 49-7841-5031
- Intl Israel, Tel Aviv**  
ISLambda Electronics Ltd., TEL: 972-3-242-194

- Intl Japan, Tokyo**  
Nemic-Lambda K.K., TEL: 03-490-5

## Linear Technology Corpo

Linear Technology Corporation  
1630 McCarthy Blvd.  
Milpitas, California 95035-7487  
408-942-0810  
TELEX: 172 110

### Distributors

- Intl Austria**  
Bacher Elektronische Gerate Gmbh  
222-835646
- Intl Belgium**  
Microtron, TEL: 32-16-60586
- Intl Denmark**  
E. V. Johansen Elektronik A/S, TI  
839033
- Intl Finland**  
OY Fintronic AB, TEL: 358-0-6926
- Intl France**  
Tekelec Airtronic, TEL: 33-1-5347
- Intl Italy**  
Eledra 3 S S.P.A., TEL: 39-2-3497
- Intl Japan**  
Japan Macnics Corporation, TEL:  
0022
- Intl Japan**  
Technology Trading Corporation,  
03-639-5581
- Intl Japan**  
Teijin Advanced Products Corpora  
044-506-4670
- Intl Netherlands**  
Alcom Electronics BV, TEL: 31-10
- Intl Norway**  
Eltron A/S, TEL: 47-2-462870
- Intl Spain**  
Amitron, S.A., TEL: 34-1-2485863
- Intl Sweden**  
Svensk Teleindustri, TEL: 46-8-33
- Intl Switzerland**  
Stolz AG, TEL: 41-56-840151
- Intl United Kingdom**  
Dialogue Distribution, Ltd., TEL: 44-682001
- Intl United Kingdom**  
Microcall, Ltd., TEL: 44-84421-54
- Intl West Germany**  
Alfred Neye, TEL: 49-4106-6121
- Intl West Germany**  
Metronik, TEL: 49-89-6114063
- Intl Yugoslavia**  
Ellyptic AG, TEL: 41-1-4931000

## LSI Computer Systems

LSI Computer Systems, Inc.  
1235 Watt Whitman Road  
Melville, New York 11747  
516-271-0400  
TWX: 510-226-7833

### Specific product information:

Ronald Colino

### Application engineering:

Ronald Colino

### Literature:

Nancy Leidig

### Price and delivery:

Nancy Leidig

### Follow-up an order:

Nancy Leidig

### All other information:

Nancy Leidig

For Intl. sales: Mavex Technology, 516-473

**IC MASTER**

**LSI Computer Systems (Cont)**

**Sales Office & Representatives**

- AL** Huntsville  
REMCO, 205-533-3763
- CA** Santa Clara  
Fischer Assocs., 408-733-7350
- CA** Westlake Village  
Western Mktg., 213-889-6270
- GA** Atlanta  
REMCO, 404-477-6937
- IL** Elmhurst  
Fiat Engrg. Assocs., 312-547-6200
- MD** Fulton  
Gans & Pugh, Inc., 301-953-3724
- NH** Merrimack  
Mech-Tron Co., 603-424-5300
- NJ** Hasbrouck Heights  
Comp-Tech Sales, 201-288-7400
- NY** New Hyde Park  
Comp-Tech Sales, 516-593-2628
- NY** Penfield  
Barthel Entprs., 716-377-3018
- NC** Raleigh  
REMCO, 919-847-5079
- OH** Cleveland  
Components, Inc., 216-243-9200
- OH** West Carrollton  
Components, Inc., 513-866-0661
- TX** Houston  
Electronic Mktg. Assoc., 713-498-8120
- VA** Arlington  
Gans & Pugh, Inc., 703-527-3262

**LSI Logic**

LSI Logic Corporation  
1601 McCarthy Boulevard  
Milpitas, California 95035  
408-263-9494  
TELEX: 172 153

**Sales Office & Representatives**

- AL** Huntsville  
Montgomery Marketing, 205-830-0498
- AZ** Scottsdale  
Luscombe Engineering, 602-949-9333, TWX 910-950-1333
- CA**  
LSI Logic Milpitas, 408-263-9494, TLX 172153
- CA** (Northern)  
Taarcom, Inc., 415-960-1550, TWX 296991
- CA** (Southern)  
Centaur Corporation, 714-556-5420, TWX 910-595-2887
- CO** Littleton  
Simpson Associates, 303-794-8381, TWX 910-935-0719
- CT** Waterbury  
HLM Associates, 203-753-9894
- FL** Altamonte Springs  
Semtronic Associates, 305-831-8233
- GA** Norcross  
Montgomery Marketing, 404-447-6127
- IL** (Southern)  
Design Solutions, 314-227-7170
- MD** Columbia  
Delta III Associates, 301-730-4700, TWX 710-862-1906
- MI** Detroit  
Rathsburg Associates, 313-882-1717
- MN** Minneapolis  
HMR, 612-831-7400

- MO** St. Louis  
Design Solutions, 314-227-7170
- NY** Rochester  
T-Squared Electronics, 716-924-9101, TWX 510-254-8542
- NC** Cary  
Montgomery Marketing, 919-467-6319, TWX 510-920-0634
- OR** Portland  
Westerberg & Associates, 503-297-1719
- TX** Dallas  
Technology Sales, 214-380-0200
- UT** Salt Lake City  
Simpson Associates, 801-566-3691
- WA** Bellevue  
Westerberg & Associates, 206-453-8881
- WI** (East)  
Oasis Sales, 312-640-1850
- WI** (West)  
HMR, 612-831-7411, TLX 029-0537, TWX 910-576-2755
- Can** B.C.  
Westerberg & Associates, 206-453-8881, TWX 910-443-2319
- Can** Ottawa, Ontario  
RFQ Limited, 613-820-8445, TWX 610-562-1973

**Master Logic**

Master Logic Corporation  
761 E. Evelyn Avenue  
Sunnyvale, California 94086  
408-732-7777

**Specific product information:**

Stephen R. Allen

**Application engineering:**

Stephen R. Allen

**Literature:**

Stephen R. Allen

**Price and delivery:**

Stephen R. Allen

**Follow-up an order:**

Stephen R. Allen

**All other information:**

Charles A. Allen

**Matrix**

Matrix Corporation  
1639 Green Street  
Raleigh, North Carolina 27603  
919-833-2837

**Matrox**

Matrox Electronic Systems Ltd.  
5800 Andover Avenue, T.M.R.  
Montreal, Quebec, H4T 1H4, Canada  
514-735-1182  
TELEX: 05-825651

**Sales Office & Representatives**

- Intl** Australia, Epping  
Mace Co. Pty. Ltd., TEL: 86-4060
- Intl** Austria, Vienna  
Bacher Elektronische Gerate, Ges. m.b.h., TEL: (0220) 835646-0
- Intl** Denmark, Glostrup  
Jorgen Andersen Ingeniørfirma A.S., TEL: (02) 92 88 88
- Intl** England, London  
Perdix Comps., TEL: 01-690 1914/5
- Intl** Finland, Helsinki  
Oy Fintronic, AB, TEL: 90-692 60 22

- Intl** France, Asnières  
Metrologie, TEL: 791 44 44
- Intl** Greece, Athens  
American Tech. Entprs. S.A., TEL: 8219.470-8811.271
- Intl** Israel, Givatayim  
EIM Intl. Elctns. Ltd., TEL: (03) 744-041
- Intl** Italy, Milano  
3-G Elctns. s.r.l., TEL: 5455951
- Intl** Japan, Tokyo  
Internix Inc., TEL: 369-1101
- Intl** Netherlands, Drimmolen  
Matrox Benelux, TEL: 0 1626-3850
- Intl** New Zealand, Auckland  
GTS Engrg., Ltd., TEL: 546-745
- Intl** Norway, Oslo  
Nordisk Elektronik AS, TEL: 02-55-2485
- Intl** Portugal, Lisboa  
Ditram Components E Electronics Lta
- Intl** Spain, Madrid  
Unitronics S.A., TEL: 242-52-04
- Intl** Sweden, Partille  
Micro Scandia Industriautomation AB, TEL: 32685
- Intl** Switzerland, Zurich  
Relais-Rohr, TEL: (01) 301 09 70
- Intl** Taiwan, Taipei  
Multitech Intl. Corp., TEL: (02) 7681232
- Intl** West Germany, Munchen  
Ernst Rauscher Systemberatung, TEL: 089-878061

**Maxim Integrated Products**

Maxim Integrated Products  
510 N. Pastoria  
Sunnyvale, California 94086  
408-737-7600

**Micrel**

Micrel  
1235 Midas Way  
Sunnyvale, California 94086  
408-245-2500  
TWX: 910-379-0007

**MCE**

MCE Semiconductor  
1111 Fairfield Drive  
W. Palm Beach, Florida 33407  
305-845-2837  
TELEX: 513463 (MCE NPAB)

**Sales Office & Representatives**

- Intl** United Kingdom, Tewkesbury  
MCE, Ltd., TEL: 0684-297777
- Intl** West Germany, Munich  
MCE Micro-Circuit Engrg. Vertr. GmbH, TEL: 089-463085
- Intl** West Germany, Nuernberg  
MCE Micro-Circuit Engrg. Vertr. GmbH, TEL: 0911-533933

**Microcircuits Technology**

Microcircuits Technology, Inc.  
1157 San Antonio Road  
Mountain View, California 94043  
415-969-3600  
TELEX: 172 750

# MANUFACTURERS & DISTRIBUTORS DIRECTORY

MANUFACTURERS & DIST DIRECTORY

## Micro Computer Control

Micro Computer Control  
P.O. Box 275  
Hopewell, New Jersey 08525  
609-466-1751

## Micro Innovators

Micro Innovators Inc.  
2348A Walsh Avenue  
Santa Clara, California 95051  
408-988-0911

## Micro-Link

Micro-Link Corporation  
14602 N. U.S. Hwy 31  
Carmel, Indiana 46032  
317-846-1721  
TWX: 810-260-2634

### Distributors

- intl **Doc, France**  
YREL, TEL: 956-8142
- intl **Tsoby, Sweden**  
EM Marketing AB, TEL: 46 0762 13570
- intl **Worms, W. Germany**  
GK Electronics, TEL: 062-41-5-32-99

## Micro Networks

Unitrode Corporation  
Micro Networks Company  
324 Clark Street  
Worcester, Massachusetts 01606  
617-852-5400  
TELEX: 951808  
TWX: 710-340-0067

- Specific product information:**  
Mark Vaughan ..... Ext. 219
- Application engineering:**  
Mark Vaughan ..... Ext. 219
- Literature:**  
Karen LaVigne ..... Ext. 220
- Price and delivery:**  
Linda Palmer ..... Ext. 275
- Follow-up an order:**  
Linda Palmer ..... Ext. 275
- All other information:**  
John Munn ..... Ext. 225

### Sales Office & Representatives

- intl **Australia, Berwood**  
R & D Elctns., TEL: 288-8232
- intl **Austria, Bad Vöslau**  
Kapla Electronik, TEL: 02252-7459
- intl **Belgium, Brussels**  
Belram S.A., TEL: 34.33.32
- intl **Belgium, Bruxelles**  
I.S.I., TEL: 02/660-13 56
- intl **Cyprus, Nicosia**  
Poly Elctns., TEL: 21-27982
- intl **Denmark, Copenhagen**  
Semicap, TEL: 01-221510
- intl **England, Sunbury-on-Thames**  
Pascall Elctns. Ltd., TEL: Sunbury, 09327-87418
- intl **Finland, Helsinki**  
Turion OY, TEL: 90-377787
- intl **France, Orsay**  
Microel, TEL: 6-9070824

- intl **Germany, Munchen**  
Tekelec Airtronic, TEL: (089) 594-621
- intl **Greece, Athens**  
Theodore D. Tzitzinias, TEL: 942 4071
- intl **Hong Kong, Hong Kong**  
Schmidt & Co. (H.K.) Ltd., TEL: 5-455644
- intl **India, Trivandrum**  
Kryonix, TEL: 63805
- intl **Israel, Bnei Brak**  
Electrondart Ltd., TEL: 3-782460
- intl **Italy, Milano**  
Microelit S.R.L., TEL: 46.90.444
- intl **Japan, Tokyo**  
Internix, TEL: (03) 369-1101
- intl **Korea, Seoul**  
Han Sun Intl., TEL: 28-0958
- intl **Netherlands, Breda**  
Indelec B.V., TEL: 076-142333
- intl **Norway, Oslo**  
Eltron, TEL: 02-15-20-51
- intl **South Africa, Bryanston**  
S'Electronics
- intl **Spain, Madrid**  
Compania Electronica de Technicas, TEL: 754-4530
- intl **Sweden, Vallingby**  
Scancopter, TEL: 08/38 00 65
- intl **Switzerland, Zurich**  
Telemeter AG, TEL: 01-202-7872
- intl **Taiwan, Taipei**  
Tai I Trading Co., Ltd., TEL: 393 61111

## Micropac

Micropac Industries, Inc.  
905 East Walnut Street  
Garland, Texas 75040  
214-272-3571  
TWX: 910-860-5186  
**All other information:**  
Ext. 23, 37

### Sales Office & Representatives

- intl **Austria, Wien**  
Transistor Vertriebs, TEL: 0043/222829451
- intl **Belgium, Brussels**  
Manudax, TEL: 00322/22152500
- intl **Denmark, Copenhagen**  
Semicap, TEL: 00451/221510
- intl **France, Boulogne**  
ASAP, TEL: 00331/6047878
- intl **France, Paris**  
Comptoir Commercial D'importation (CCI),  
TEL: 00331/2362045
- intl **Great Britain, Thame-Oxford**  
Gain Elctns., TEL: 0044/844215477
- intl **Great Britain, Harrold-Bedford**  
Synchro Svcs., TEL: 0044/234720575
- intl **Italy, Milano**  
El Pack S.R.L., TEL: 00392/6471673
- intl **Netherlands, Nijmegen**  
Varilec B.V., TEL: 00318/0445660
- intl **Norway, Teusbert**  
UMI A.S., TEL: 004733/22214
- intl **Sweden, Solna**  
AB Betoma, TEL: 00468/820280
- intl **Switzerland, Baden**  
Comeltec AG, TEL: 004156/223152
- intl **Switzerland, Zurich**  
Egli, Fischer Co. Ltd., TEL: 00411/2020234
- intl **West Germany, Geesthacht**  
Microlec, TEL: 04152/72324
- intl **West Germany, Homburg**  
IV Elctn., TEL: 06172/23061
- intl **West Germany, Munich**  
DIPL-Ing. Ernst Fey Nachf., TEL: 089-184041
- intl **West Germany, Munich**  
Neumueller GmbH, TEL: 089/61181

- intl **West Germany, Munich**  
Nucletron Vertriebs GmbH, TEL: 089/146081
- intl **West Germany, Stuttgart**  
Ditronic GmbH, TEL: 0711/724844

## Micro Power Systems

Micro Power Systems, Inc.  
3100 Alfred Street  
Santa Clara, California 95050  
408-727-5350  
TWX: 910-338-0154

- Specific product information:**  
Standard Products Division ..... ext. 237
- Literature:**  
Standard Products Division ..... ext. 237

### Sales Office & Representatives

- AL **Huntsville**  
R.W. Mitscher Co., Inc., 205-852-7676
- AZ **Tempe**  
Sun West Mktg. Assoc., 602-241-8163
- CA **Los Alamitos**  
Reed Elec. Mktng., 213-598-6676, 714-821-9600
- CA **San Diego**  
D $\frac{1}{2}$ 32 Sales, Inc., 619-560-6266
- CA **Sunnyvale**  
NES, 408-980-1950
- CO **Littleton**  
Quatra Inc., 303-795-3187
- FL **Clearwater Beach**  
Perrott Assocs., Inc., 813-443-5214
- FL **Orlando**  
Perrott Assocs., Inc., 305-298-7748
- GA **Lawrenceville**  
R.W. Mitscher Co., Inc., 404-923-3239
- IL **Arlington Heights**  
Coombs Assoc., 312-439-9810
- IL **Rolling Meadows**  
KMA Sales, 312-398-5300
- IN **Ft. Wayne**  
Coombs Assocs., 219-747-7661
- IA **Marion**  
REP Assocs., 319-373-0152
- KS **Parler**  
S.W. Woolard Co., 913-898-6552
- KY **Louisville**  
K.W. Elctrn. Sales, 502-451-1860
- MD **Baltimore**  
Stemler Assocs., 301-944-8262
- MA **Berlinton**  
Dynamic Sales of New England, 617-272-5676
- MI **Livonia**  
Lowell Wendt Mktg. Co., 313-464-2722
- MI **Livonia**  
Reptron Electronic, Inc., 313-525-2700
- MN **Minneapolis**  
Peterson Sales Assocs., 612-884-3483
- MO **Rockville**  
Stemler Associates, 301-774-2268
- NJ **Freehold**  
Comtronic Assoc., Inc., 201-431-3374
- NJ **Marlton**  
Stemler Assocs., 609-966-4070
- NY **Bainbridge**  
R.W. Mitscher Company, 607-967-7192
- NY **Buffalo**  
R.W. Mitscher Co., Inc., 716-633-7970
- NY **Melville**  
Comtronic Assocs., 516-249-0505
- NY **N. Syracuse**  
R.W. Mitscher Co., Inc., 716-586-4012
- NY **Plainview**  
Comtronic Assoc., Inc., 516-681-7936
- NY **Red Hook**  
R.W. Mitscher Co., Inc., 914-758-8375



# IC MASTER

## Micro Power Systems (Cont)

**NY Rochester**  
R.W. Mitscher Co., Inc., 716-227-0839

**NY Rockville**  
Comtronic Assoc., Inc., 516-536-4020

**NY Staten Island**  
Comtronic Assoc., Inc., 212-979-3973

**NY Whitestone**  
Comtronic Assoc., Inc., 212-767-7461

**NC Raleigh**  
R.W. Mitscher Company, 919-876-0160

**OH Dayton**  
K.W. Elctrn. Sales, Inc., 513-890-2150

**OH Shaker Heights**  
K.W. Elctrn. Sales, Inc., 216-491-9177

**OH Worthington**  
K.W. Electronic Sales, Inc., 614-888-0483

**OK Tulsa**  
Quality Component, 918-664-8812

**OR Tigard**  
Elctrn. Comp. Sales, 503-245-2352

**PA Allison Park**  
K.W. Electronic Sales, Inc., 412-487-4300

**TN Johnson City**  
R.W. Mitscher Co., Inc., 615-282-6240

**TN Johnson City**  
R.W. Mitscher Company, 615-282-6240

**TX Austin**  
TMI, 512-835-0064

**TX Carrollton**  
TMI, 214-387-3601

**TX Houston**  
TMI, 713-777-9228

**WA Mercer Island**  
Elctrn. Comp. Sales, 206-232-9301

**WI Milwaukee**  
KMA Sales, 414-259-1771

### Distributors

**CA Laguna Hills**  
MP Systems, 714-770-6411

**CA Redwood**  
Jan Devices, 213-708-1100

**CA Santa Ana**  
Pacesetter Elctns., 714-557-7131

**CA Sunnyvale**  
Future Elctns., 408-745-6570

**CA Sunnyvale**  
Pacesetter Elctns., 408-734-5470

**CO Wheatridge**  
Bell Inds., 303-424-1985

**CT Danbury**  
Pilgrim Elctns., 203-792-7274

**FL Clearwater**  
Diplomat Elctns., 813-443-4515

**FL Orlando**  
Chip Supply Company, 305-275-3810

**GA Atlanta**  
Wholesale Industrial Elect., 404-447-8431

**IL Elk Grove Village**  
GBL Gould Elctns., 312-593-3220

**MD Baltimore**  
Whitney Dists., 301-944-8080

**MA Wilmington**  
Aved Elctns., Inc., 617-657-8310

**MN Minneapolis**  
The Joel Co., 612-545-5669

**NM Albuquerque**  
Alliance Elec., 505-292-3360

**NY Buffalo**  
Summitt Dists., 716-887-2800

**NY E. Syracuse**  
Add Elctns., 315-437-0300

**OH Englewood**  
Argent Inds., 513-836-8633

**OH Solon**  
REH Elctns., 216-248-1636

**SC Columbia**  
R.W. Mitscher Company, 803-736-4544

**TX Addison**  
Quality Comps., Inc., 214-387-4949

**TX Austin**  
Quality Comps., Inc., 512-835-0220

**TX Houston**  
Quality Comps., Inc., 713-772-7100

**UT Salt Lake City**  
Bell Industries, 303-795-3187

**WA Seattle**  
Robert E. Priebe Co., 206-682-8242

**WI Maquon**  
Taylor Elctrn. Co., 414-241-4321

**Intl Australia, Rozelle**  
Ampec Elctns., TEL: (02) 818-1166

**Intl Austria, Vienna**  
NANO-80, TEL: (0222) 387638

**Intl Denmark, Copenhagen**  
Advanced Elctns. of Denmark Aps, TEL: 011-45-2-194433

**Intl England**  
B.A. Electronics

**Intl England**  
Coole Marketing Services, Ltd., TEL: 011-44-734-700453

**Intl England**  
MDL Components, Ltd.

**Intl England**  
Quarndon Electronics Semiconductor, TEL: 011-332-363-291

**Intl England, Essex**  
Pronto Elctrn. Sys., Ltd., TEL: 01-599 3041/2

**Intl Finland, Helsinki**  
Micronas, TEL: 011-358-0-5623-977

**Intl France, Gentilly**  
Radio Television Francaise, TEL: 011-33-1-664-1101

**Intl France, La Chesnay**  
A2M, TEL: 011-3-955-32-49

**Intl France, Paris**  
Unirep, TEL: 011-33-1-686-3942

**Intl Holland, Zoetermeer**  
Tekelec Airtronic, TEL: 079-310100

**Intl Hong Kong, Singapore**  
Sentry Tech., TEL: telex: 786-38646

**Intl Israel, Tel Aviv**  
STG Intl. Ltd., TEL: (03) 248231

**Intl Italy, Fiorentino**  
Intelco, TEL: telex: 843-573016

**Intl Italy, Milano**  
Tekelec Airtronic, TEL: 02-7380641

**Intl Italy, Torino**  
Inter-Rep, TEL: 221-422-843

**Intl Japan, Tokyo**  
Kaigai Elctns., TEL: 586-0471

**Intl Nederland, Haarlem**  
Techmaton, TEL: 04189-2222

**Intl Norway, Hvalstad**  
Estronic, TEL: 02-786010

**Intl South Africa, Ransburg**  
South Continental Devices, TEL: (011) 798-2400

**Intl Spain, Barcelona**  
Diode Espana, TEL: 217 26 19-217-26-52

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Kimates Iberica, TEL: 230-0204, 230-0205

**Intl Spain, Bilbao**  
Diode Espana, TEL: 424 86 55-423-65-83

**Intl Spain, Madrid**  
Diode Espana, TEL: 455 36 86

**Intl Sweden, Danderyd**  
Setron AB, TEL: 011-0-753-0055

**Intl Switzerland, Schweiz**  
Anatec AG, TEL: (042) 367633

**Intl West Germany, Hamburg**  
Microsan, TEL: 040-630-5067

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Adicom, TEL: 011-49-89-479699

**Intl West Germany, Munich**  
Kontron Elektronik GmbH, TEL: 011-49-89-31901-381

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## Micro Sciences

Micro Sciences Corporation  
145 Commack Road  
Commack, New York 11725  
516-499-0635

## MillerTronics

MillerTronics  
303 Airport Road  
Greenville, South Carolina 29607  
803-242-9232

## Miller Technology

Miller Technology Inc.  
647 North Santa Cruz Avenue  
Los Gatos, California 95030  
408-395-2032

## Mitel Corp.

**Mitel Semiconductor**  
P.O. Box 13320  
Kanata, Ontario, Canada K2K 1X5  
613-502-5630  
TELEX: 053-3221  
TWX: 610-562-9529

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Head Office

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Customer Service

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**All other information:**  
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Pro Assocs., 408-358-3861

**CA San Diego**  
L & S Assocs., 619-455-0055

**CA San Diego**  
Mitel Semiconductor, 619-276-3421

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Mitel Semiconductor, 408-249-2111

**CO Denver**  
D/Z Associates, 303-429-9369

**CT Kensington**  
Norm Stachelek, 203-828-6161

**FL Boca Raton**  
Mitel Semiconductor, 305-994-8588

**FL Clearwater**  
Hutto, Hawkins, Peregoy (H.H.P.), 813-797-4447

**FL Ft. Lauderdale**  
Hutto, Hawkins, Peregoy (H.H.P.), 305-971-5750

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MA	<b>Gurtington</b> Access Systems, 617-229-2433	IL
MI	<b>Troy</b> Scott Electrs. Inc., 513-791-2513	IN
MN	<b>New Brighton</b> T.W.C. Sales, 612-636-1770	MA
MO	<b>St. Louis</b> Advanced Technical Sales, 314-721-4401	MN
NH	<b>Amherst</b> Access Systems, 603-673-0554	MS
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NY	<b>Endwell</b> Tri Tech Electrs., Inc., 607-754-1094	OH
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NY	<b>Fishkill</b> Tri Tech Electronics, Inc., 914-897-5611	TX
NY	<b>Valley Cottage</b> L.R.C. Assocs., 914-268-4435	TX
NC	<b>Raleigh</b> A.B. Andrews & Co., 919-549-0563	TX
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OH	<b>Cleveland</b> Scott Electrs. Inc., 216-473-5050	WA
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TX	<b>Richardson</b> Logic 1, 214-234-0765	Can
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Can	<b>Richmond, British Columbia</b> Deskin Sales, 604-273-4963	Can
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AZ	<b>Tempe</b> Anthem Electronics, 602-244-0900	Can
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CA	<b>Santa Ana</b> Pacesetter Electrs., 714-557-7131	Can
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CA	<b>Tustin</b> Anthem Electronics, 714-730-8000	Can
CO	<b>Englewood</b> Anthem Electronics, 303-790-4500	Can
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CO	<b>Westminster</b> D-Z Associates, 303-429-9369	Can
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NY	<b>Piscataway</b> ACI Electrs. Corp., 516-293-6630	Can
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TX	<b>Addison</b> Quality Comps. Inc., 214-387-4949	Can
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TX	<b>Houston</b> Quality Comps. Inc., 713-722-7100	Can
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WA	<b>Redmond</b> Pacom Inc., 206-883-9200	Can
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Can	<b>Burnaby, British Columbia</b> R-A-E Indl. Electrs. Ltd., 604-291-8866	Can
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Can	<b>Edmonton, Alberta</b> R-A-E Indl. Electronics Ltd., 403-451-0001	Can
Can	<b>Ottawa, Ontario</b> Zentronics, 613-238-6411	Can
Can	<b>Richmond, British Columbia</b> Zentronics, 604-273-5575	Can
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Can	<b>St. Laurent, Quebec</b> Zentronics, 514-735-5361	Can
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Can	<b>Waterloo, Ontario</b> Zentronics, 519-884-5700	Can
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- Int'l Denmark**  
Mitel Semiconductor, TEL: 01-134712
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Cybernetics, TEL: 01030 1 958 8464
- Int'l Holland, Wassenaar**  
Werner Elctrns. BV, TEL: 01751 19225
- Int'l Hong Kong, Kowloon**  
Draco Intl Ltd., TEL: 3-891331, 3-446117
- Int'l Hong Kong, Tsuen Wan**  
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- Int'l India, Bangalore**  
Flash Entprs., TEL: 72660
- Int'l Italy, Milano**  
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- Int'l Japan, Tokyo**  
Mitel Japan, TEL: (03) 402-9941
- Int'l Japan, Tokyo**  
Tokyo Electron Ltd., TEL: 03-343-4411
- Int'l Norway, Asker**  
Intelco AS, TEL: (47 2) 78 06 76
- Int'l Switzerland, Zurich**  
EGLI Fischer & Co., TEL: (01) 202 02 34
- Int'l United Kingdom, Essex**  
Pronto Elctrns. Sys., TEL: 01-554-6222

**Mitsubishi Electronics**

Mitsubishi Electronics America, Inc.  
Semiconductor Division  
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Sunnyvale, California 94086  
408-730-5900  
TELEX: 172296 MELA SUVL.  
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**DIPLOMAT ELECTRONICS CORPORATION**

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Monolithic Memories, Inc.  
2175 Mission College Blvd.  
Santa Clara, CA 95050  
408-970-9700

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Summit Sales, 602-998-4850
- CA Auburn**  
Thresum Assoc., 916-269-1543.
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Monolithic Memories, 714-556-1216
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Monolithic Memories, 408-996-1477
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Thresum Assocs., 408-996-9889
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- CA Irvine**  
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Monolithic Memories, 714-556-1216
- CA San Diego**  
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- FL Clearwater**  
Dyne-A-Mark, 813-441-4702
- FL Fort Lauderdale**  
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- FL Palm Bay**  
Dyne-A-Mark, 305-727-0192
- GA Norcross**  
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- GA Tucker**  
Rep, Inc., 404-938-4358
- IL Lombard**  
Monolithic Memories, 312-932-1940
- IL Naperville**  
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- IL Rolling Meadows**  
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Monolithic Sales, 301-296-2444
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Monolithic Memories, 617-875-7373
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- MI Edina**  
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- MO Ballwin**  
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- NJ Sussex**  
Monolithic Memories, 201-875-9430
- NJ Teaneck**  
R. T. Reid Assoc., 201-692-0200
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- NY E. Rochester**  
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- NY Endwell**  
Tri-Tech Electronics, 607-754-1094
- NY Fayetteville**  
Tri-Tech Electronics, 315-446-2881
- NY Fishkill**  
Tri-Tech Electronics, 914-897-5611
- NC Raleigh**  
Rep, Inc., 919-851-3007
- OH Cincinnati**  
Makin Assocs., 513-871-2424
- OH Columbus**  
Makin Assocs., 614-459-2423
- OH Dayton**  
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Makin Assocs., 216-921-0080
- OK Tulsa**  
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- PA Oroland**  
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- TX Dallas**  
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West Assocs., 713-777-4108
- UT Salt Lake City**  
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Can	<b>Brampton, Ontario</b> Cantec, 416-791-5922
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Can	<b>Ottawa, Ontario</b> Cantec, 613-725-3704
Can	<b>Waterloo, Ontario</b> Cantec, 519-744-6341
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AZ	<b>Tempe</b> Anthem Elctns., 602-244-0900
AZ	<b>Tempe</b> Bell Inds., 602-966-7800
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AZ	<b>Tucson</b> Kierulff Elctns., 602-624-9986
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CA	<b>Chatsworth</b> Anthem Elctns., 213-700-1000
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CA	<b>Los Angeles</b> Kierulff Elctns., 213-725-0325
CA	<b>Newport Beach</b> Arrow Elctns., 714-851-8961
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CA	<b>San Diego</b> Anthem Elctns., 619-279-5200
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CA	<b>San Jose</b> Anthem Elctns., 408-946-8000
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CA	<b>Tustin</b> image Elctns., 714-730-0303
CA	<b>Tustin</b> Kierulff Elctns., 714-731-5711
CO	<b>Aurora</b> Arrow Elctns., 303-696-1111
CO	<b>Englewood</b> Anthem Electronics, 303-790-4500
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CO	<b>Wheatridge</b> Bell Inds., 303-424-1985
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CT	<b>Wallingford</b> Arrow Elctns., 203-265-7741
CT	<b>Wallingford</b> Kierulff Elctns., 203-265-1115
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GA	<b>Norcross</b> Hall-Mark Elctns., 404-447-8000
GA	<b>Norcross</b> Kierulff Elctns., 404-447-5252
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**TX Austin**  
A.C.T. Electronics, 512-452-5254

**TX Dallas**  
RM Elctns., 214-263-8361

**TX Houston**  
A.C.T. Electronics, 713-496-4000

**UT Salt Lake City**  
Act Electronics, 801-487-8131

**WA Kent**  
Shannon Ltd., 206-763-0545

**WA Seattle**  
Zepher Electronics, 206-241-1980

**WI New Berlin**  
RM Elctns., 414-784-4420

**Oliver Advanced Engineering**

Oliver Advanced Engineering, Inc.  
676 West Wilson Avenue  
Glendale, California 91203  
213-240-0080  
TELEX: 194773  
Specific product information:  
Marketing & Sales Department ..... 213-240-0080

**Sales Office & Representatives**

**AZ Mesa**  
RLS Instruments, 602-892-9329

**CA Carson**  
AMASCO, 213-516-8561

**CA Sunnyvale**  
AMASCO, 408-733-8690

**IL Wheeling**  
Engrg. Prods. Assocs., Inc., 312-537-4860

**IN Indianapolis**  
Engrg. Prods. Assocs., Inc., 317-894-7156

**IA Iowa City**  
Engrg. Prods. Assocs., Inc., 319-354-5344

**LA Louisiana**  
Industrial Digital Systems, 504-899-1654

**MA Medford**  
Comark, 617-359-8161

**MN Minneapolis**  
Engrg. Prods. Assocs., Inc., 612-925-1883

**NM Albuquerque**  
RLS Instruments, 505-265-4187

**TX Houston**  
Industrial Digital Systems, 713-988-9421

**WA Kirkland**  
Electronic Marketing Associates, 206-827-9350

**Intl Austria, Wien**  
Elbatex Ges.m.b.H., TEL: (0222) 86 56 11 Serie

**Intl Belgium, Zentloeuw**  
Belcomp PVBA, TEL: 011/78 05 12

**Intl Brazil, Sao Paulo**  
Filcres Instruments, TEL: 011-223-7388

**Intl Denmark, Hershøim**  
C-88 aps, TEL: 02\* 57 08 88

**Intl England, Alton, Hampshire**  
MTL Microsystems, TEL: 0420-88022

**Intl France, Orsay**  
Microel "Le Parana", TEL: 907-08-24

**Intl Germany, Munich**  
Scantec GmbH, TEL: (089) 134093

**Intl India, Bombay**  
Echbee Corp., TEL: 25 34 89/25 83 41

**Omnibyte**

Omnibyte Corporation  
245 W. Roosevelt Road, Building 1-5  
West Chicago, Illinois 60185  
312-231-6880

**Distributors**

**Intl Almere-Haven, Holland**  
DeGids & Feldman, TEL: 730953

**Intl Barcelona, Spain**  
Lana Sarrate, S.A., TEL: 204-44-50

**Intl Berkshire, England**  
Measurement Systems Ltd., TEL: (44) 0635 45420

**Intl Christchurch, New Zealand**  
E. C. Gough Ltd., TEL: (03) 798-740

**Intl Hyderabad, India**  
International Trade Agencies, Ltd., TEL: 31543-31216

**Intl Milano, Italy**  
Auriema Italia, TEL: 430, 602

**Intl Solna, Sweden**  
Master Data AB, TEL: 08-83 01 55

**Intl Stuttgart, W. Germany**  
Claus Polack, TEL: Telex 2627-7111038

**Intl Taiwan, Taipei**  
Strong Electronics Co., Ltd., TEL: (02) 521-8826-7

**Intl Victoria, Australia**  
Rank Electronics Pty. Ltd., TEL: (03) 29 3724

**Intl Zurich, Switzerland**  
Omni Ray, TEL: (01) 47 8200

**Onset**

Onset Computer Corporation  
199 Main Street, P.O. Box 1016  
North Falmouth, Massachusetts 02556  
617-563-2267

**Optical Electronics**

Optical Electronics Inc.  
P.O. Box 11140  
Tucson, Arizona 85734  
602-624-8358  
TWX: 910-952-1283

**Panasonic**

Panasonic Industrial Company/Electronic Components Division  
Matsushita Electric Corporation of America  
1 Panasonic Way  
Secaucus, New Jersey 07094  
201-348-5269  
TWX: 710-992-8920

**Sales Office & Representatives**

**AL Huntsville**  
REP, Inc., 205-881-9270

**AZ Phoenix**  
Stitzer-Dodd, Inc., 602-944-2237

**CA Newport Beach**  
West, Inc., 714-751-5656

**CA Pleasanton**  
Moulthrop Sales, Inc., 415-864-0550

# MANUFACTURERS & DISTRIBUTORS DIRECTORY

MANUFACTURERS & DIST DIRECTORY

## Panasonic (Cont)

**CO** **Denver**  
Component Sales, 303-759-1666

**CT** **Madison**  
Conti-Younger Associates, Inc., 203-245-1136

**FL** **Coral Springs**  
CBC Electronics, 305-755-1111

**FL** **Longwood**  
CBC Electronics, 305-831-5380

**GA** **Tucker**  
REP, Inc., 404-938-4358

**IL** **Chicago**  
Brock & Cushman, Inc., 312-622-4110

**IN** **Carmel**  
Marketing Engineers, Inc., 317-872-5665

**KS** **Olathe**  
Rush & West Inc., 913-764-2700

**MD** **Ellicott City**  
Marketing Technology, 301-465-5944

**MA** **Burlington**  
Conti-Younger Associates, Inc., 617-273-1582

**MI** **Brighton**  
Electronic Sources, Inc., 313-227-3598

**MN** **Minneapolis**  
Electro Mark, Inc., 612-881-3306

**MO** **Bellwin**  
Rush & West Inc., 314-394-7271

**NJ** **Hackensack**  
Meg Electronic Sales, 201-489-7700

**NJ** **Mount Laurel**  
Comtek, Inc., 609-235-8505

**NY** **North Syracuse**  
Advanced Components Corp., 315-669-2671

**NC** **Raleigh**  
REP, Inc., 919-851-3007

**OH** **Cleveland**  
Del Steffen & Associates, 216-461-8333

**OH** **Dayton**  
Del Steffen & Associates, 513-293-3145

**OH** **Lexington**  
Del Steffen & Associates, 419-884-2313

**OR** **Tigard**  
Electronic Sources, 503-684-0040

**PA** **Pittsburgh**  
Del Steffen & Associates, 412-276-7366

**TN** **Jefferson City**  
REP, Inc., 615-475-4105

**TX** **Austin**  
Dunbar Associates, 512-346-9337

**TX** **Dallas**  
Dunbar Associates, 214-239-7151

**UT** **Salt Lake City**  
Component Sales, 801-466-8623

**VA** **Lynchburg**  
Marketing Technology, 804-384-6459

**WA** **Bellevue**  
Electronic Sources, 206-451-3500

**WI** **Monomonee Falls**  
Brock & Cushman, Inc., 414-255-2166

**Can** **Downsview, Ontario**  
Astec Components, Ltd., 416-630-8463

### Distributors

**AL** **Huntsville**  
Resistacap, Inc., 205-883-4270

**AZ** **Phoenix**  
Sterling Electronics, 602-258-4531

**AZ** **Tempe**  
Circuit Specialists, 602-966-0764

**AZ** **Tempe**  
Shelley Electronics, 602-894-2058

**CA** **Burbank**  
TAW Electronics, 213-846-3911

**CA** **Chatsworth**  
Shelley Electronics, 213-998-3333

**CA** **Hayward**  
Allied Electronics, 415-487-5250

**CA** **Irvine**  
Ratel Electronics, 714-549-8611

**CA** **Los Angeles**  
Western Electromotive, 213-820-3777

**CA** **Mt. View**  
Shelley Electronics, 415-969-1820

**CA** **San Diego**  
Shelley Electronics, 619-453-7133

**CA** **San Diego**  
Tauber Electronics Inc., 619-274-7242

**CA** **San Diego**  
Western Electromotive, 619-271-4511

**CA** **San Jose**  
Ratel Electronics, 408-946-4700

**CA** **Santa Clara**  
Western Electromotive, 408-727-7411

**CA** **Sunnyvale**  
CapSCO Sales Inc., 408-734-3020

**CO** **Boulder**  
Shelley Electronics, 303-442-2718

**CO** **Denver**  
Denver/Wintronics, 303-934-2406

**CO** **Denver**  
Integrated Electronics, 303-292-6121

**CT** **E. Haven**  
J.V. Electronics, 203-469-2321

**CT** **Hamden**  
Impact Sales Co., Inc., 203-281-1331

**FL** **Longwood**  
Interlec Components, 305-834-7777

**FL** **Miami**  
Electronic Equipment, 305-871-3500

**GA** **Augusta**  
Dixie Electronic Inc., 404-772-2055

**GA** **Cumming**  
Pan American Electronics, 404-577-2114

**IL** **Chicago**  
Prehler, Electronics Inc., 312-384-6100, 800-621-4177

**IL** **Elgin**  
Allied Electronics, 312-697-8200

**IL** **Palatine**  
Ohm Electronics, 312-359-5500

**IN** **Fort Wayne**  
Graham Electronic Supply, 219-423-3422

**IN** **Indianapolis**  
Graham Electronics, 317-634-8202

**IA** **Cedar Rapids**  
Deeco, Inc., 319-365-7551

**LA** **Metairie**  
Sterling Electronics, 504-887-7610

**MD** **Timonium**  
Zebra Electronics, 301-252-6576

**MA** **Canton**  
Impact Sales, 617-821-1500

**MA** **Wendon**  
Add Electronics, 617-478-4200

**MA** **Wilmington**  
R.C. Components Inc., 617-273-1860

**MI** **Farmington**  
Prehler Electronics Inc., 313-473-7200

**MN** **Minneapolis**  
Industrial Components Inc., 612-831-2666

**MN** **Thief River Falls**  
Digi-Key Corp., 218-681-6674

**MO** **St. Louis**  
Olive Electronics Inc., 314-426-4500

**NJ** **Bellmawr**  
Mid Atlantic Corp., 609-931-5303

**NJ** **Plainfield**  
Mars Electronics, 201-757-4332

**NJ** **Pompton Plains**  
Zebra Electronics, 201-839-9040

**NM** **Albuquerque**  
Sterling Electronics, 505-884-1900

**NY** **Bohemia**  
Zebra Electronics, 516-563-2160

**NY** **Buffalo**  
Summit Distributors, 716-887-2800

**NY** **E. Syracuse**  
Add Electronics, 315-437-0300

**NY** **Rochester**  
Summit Distributors, 716-334-8110

**NY** **Rome**  
Rome Electronics, 315-337-5400

**NC** **Charlotte**  
Dixie Electronics, 704-377-5413

**NC** **Winston-Salem**  
Dixie Electronics, 919-724-5961

**OH** **Cincinnati**  
Hughes Peters Inc., 513-351-2000

**OH** **Columbus**  
Hughes Peters Inc., 614-294-5351

**OH** **Middleburgh Heights**  
Prehler Electronics Inc., 216-243-5510

**OK** **Tulsa**  
Sterling Electronics, 918-663-2410

**OR** **Beaverton**  
Almac Electronics, 503-641-9070

**PA** **Harrisburg**  
Advacom, Inc., 717-545-5674

**PA** **McKean**  
Advacom, Inc., 814-476-7774

**PA** **Philadelphia**  
Philadelphia Electronics, 215-568-7400

**SC** **Charleston**  
Dixie Electronics, 803-554-7515

**SC** **Columbia**  
Dixie Electronics, 803-779-5332

**SC** **Florence**  
Dixie Electronics, 803-669-8201

**SC** **Greenville**  
Dixie Electronics, 803-297-1430

**TN** **Gray**  
Dixie Electronics, 615-477-3838

**TX** **Austin**  
Sterling Electronics, 512-836-1341

**TX** **Dallas**  
Computer Components Corp., 214-239-0271

**TX** **Dallas**  
Sterling Electronics, 214-243-1600

**TX** **Fl. Worth**  
Allied Electronics, 817-336-5401

**TX** **Houston**  
Sterling Electronics, 713-627-9800

**UT** **Midvale**  
Ratel Electronics, 801-562-2600

**UT** **Salt Lake City**  
Standard Supply, 801-486-3371

**VA** **Richmond**  
Sterling Electronics, 804-226-2190

**WA** **Bellevue**  
Almac Electronics, 206-643-9992

**WA** **Seattle**  
Zepher Industries Inc., 206-241-1980

**WI** **Mequon**  
Taylor Electric, 414-241-4321

**Can** **Mississauga, Ontario**  
Prelo Electronics Ltd., 416-678-0401

**Can** **Montreal, Quebec**  
Electronic Wholesalers Ltd., 514-769-8851

**Can** **Montreal, Quebec**  
Prelo Electronics Ltd., 514-389-8051

**Can** **Ottawa, Ontario**  
Electronic Wholesalers Ltd., 613-746-4413

**Can** **Ottawa, Ontario**  
Prelo Electronics Ltd., 613-226-3491

### Pico Design

Pico Design, Inc.  
1333 Lawrence Expressway, Suite 340  
Santa Clara, California 95051  
408-249-8090

**IC MASTER**

**Plessey Solid State**

**Plessey Solid State**

3 **Whitney**  
Irvine, California 92714  
714-540-9979  
TELEX: 595-1930

**Application engineering:**  
ext. 33

**Literature:**  
Burma Tayengco

**Price and delivery:**  
Pat Dilley ..... ext.49

**Follow-up on order:**  
Pat Dilley ..... ext. 27

**Sales Office & Representatives**

- AL** **Huntsville**  
R.W. Mitscher Co., Inc., 205-882-6050
- AZ** **Scottsdale**  
Thorsen Desert States, Inc., 602-998-2444
- CA** **Irvine**  
National Sales Office, 714-540-9979
- CA** **Irvine**  
Southwest District Sales Office, 714-540-9979
- CA** **Irvine**  
West Regional Sales Office, 714-540-9979
- CA** **Santa Clara**  
Logan Sales, 408-496-0660
- CA** **Sunnyvale**  
Northwest District Sales Office, 408-730-1111
- CA** **Thousand Oaks**  
S.C. Cubed, 805-496-7307
- CA** **Tustin**  
S.C. Cubed, 714-731-9206
- CO** **Englewood**  
Dynatech, Inc., 303-773-2830
- FL** **Boca Raton**  
Lawrence Associates, 305-368-7373
- FL** **Clearwater**  
Lawrence Associates, 813-584-8110
- FL** **W. Melbourne**  
Lawrence Associates, 305-724-8294
- GA** **Lawrenceville**  
R.W. Mitscher Co., Inc., 404-923-3239
- GA** **Norcross**  
Southeast Regional Sales Office, 404-447-6910
- IL** **Arlington Heights**  
Micro Sales Inc., 312-956-1000
- IL** **Arlington Heights**  
Midwest Regional Sales Office, 312-437-1860
- IN** **Ft. Wayne**  
Central District Sales Office, 219-483-4863
- IN** **Indianapolis**  
SAI Marketing Corp., 317-241-9276
- KS** **Merriam**  
Midwest Tech. Sales, 913-236-8557
- MD** **Beltsville**  
Applied Engineering Con., 301-595-5393
- MA** **Framingham**  
Eastern District Sales Office, 617-872-2288
- MI** **Brighton**  
SAI Marketing Corp., 313-227-1786
- MN** **Bloomington**  
Electronics Sales Agency, 612-884-8291
- NJ** **Marlton**  
B.G.R. Associates, 609-428-2440
- NY** **Hauppauge**  
Eastern Regional Sales Office, 516-273-3060
- NY** **Hicksville**  
Lorac Sales, 516-681-8746
- NY** **Skanateles**  
Robtron, Inc., 315-685-5731
- NC** **Charlotte**  
R.W. Mitscher Co., Inc., 714-332-8490
- NC** **Raleigh**  
R.W. Mitscher Co., Inc., 919-876-0160

- OH** **Centerville**  
SAI Marketing Corp., 513-435-3181
- OH** **Shaker Heights**  
SAI Marketing Corp., 216-751-3633
- OH** **Zanesville**  
SAI Marketing Corp.
- TN** **Johnson City**  
R.W. Mitscher Co., Inc., 615-282-6240
- TX** **Arlington**  
W. Pat Fralia Co., Inc., 817-640-9101
- TX** **Austin**  
W. Pat Fralia Co., Inc., 512-835-7911
- TX** **Houston**  
W. Pat Fralia Co., Inc., 713-772-1572
- UT** **Bountiful**  
Anderson Associates, 801-292-8991
- WI** **Brookfield**  
Micro Sales Inc., 414-782-1171
- Can** **Boiton, Ontario**  
MacKay Associates, 416-857-4302
- Can** **Burnaby**  
RAE Industrial Electronics, 604-291-8866

**Distributors**

- Can** **Mississauga, Ontario**  
GEC Canada Ltd., 416-624-8300

**Polycore Electronics**

Polycore Electronics Inc.  
1107 Tourmaline Drive  
Newbury Park, California 91320  
805-498-8832; 213-991-1061

**Precision Monolithics**

**Precision Monolithics Inc.**  
1500 Space Park Drive  
Santa Clara, California 95050  
408-727-9222

**TWX: 910-338-0528**

**Specific product information:**

ext. 162

**Application engineering:**

ext. 162

**Literature:**

ext. 186

**Price and delivery:**

ext. 171

**Follow-up on order:**

ext. 171

**All other information:**

ext. 162

**Sales Office & Representatives**

- AL** **Huntsville**  
EMA, 205-830-4030
- AL** **Huntsville**  
Pioneer/Harvey Electronics, 205-837-9300
- AZ** **Scottsdale**  
PMI Sales Office, 602-941-1946
- AZ** **Tempe**  
Anthem Electronics, 602-244-0900
- AZ** **Tempe**  
Bell Industries, 602-966-7800
- CA** **Los Angeles**  
PMI Sales Office, 213-642-0142
- CA** **San Diego**  
Bell Industries, 619-268-1277
- CA** **Santa Clara**  
PMI Sales Office, 408-727-6616
- CO** **Littleton**  
PMI Sales Office, 303-979-8533
- CO** **Wheatridge**  
Bell Industries, 303-424-1985

- CT** **Unionville**  
PMI Sales Office, 203-673-9995
- FL** **Longwood**  
PMI Sales Office, 305-331-5133
- FL** **Orlando**  
Pioneer/Harvey Electronics, 305-834-9090
- FL** **St. Petersburg**  
Hall-Mark Electronics, 813-576-8691
- GA** **Atlanta**  
EMA, 404-329-0530
- GA** **Gaithersburg**  
Pioneer/Harvey Electronics, 301-921-0660
- GA** **Norcross**  
Hall-Mark Electronics, 404-447-8000
- GA** **Norcross**  
Pioneer/Harvey Electronics, 404-448-1711
- IL** **Schaumburg**  
PMI Sales Office, 312-885-8440
- IN** **Indianapolis**  
PMI Sales Office, 800-323-8755
- KS** **Merriam**  
PMI Sales Office, 800-323-8755
- MD** **Lutherville**  
PMI Sales Office, 301-882-9222
- MA** **Natick**  
PMI Sales Office, 617-655-8900
- MA** **Westborough**  
Future Electronics, 617-366-2400
- MI** **Brighton**  
PMI Sales Office, 800-323-8755
- MN** **Bloomington**  
PMI Sales Office, 800-323-8755
- MN** **Minneapolis**  
Merit Electronics Corp., 612-546-5383
- MN** **Minnetonka**  
Pioneer/Harvey Electronics, 612-935-5444
- NJ** **Cherry Hill**  
Hall-Mark Electronics, 609-424-7300
- NJ** **Fairfield**  
Hall-Mark Electronics, 201-575-4415
- NY** **East Syracuse**  
L-MAR Assocs., Inc., 315-437-7779
- NY** **Hicksville**  
J-Square Mktg., Inc., 516-935-3200
- NY** **Rochester**  
L-MAR Assocs., Inc., 716-323-1000
- NC** **Charlotte**  
Pioneer/Harvey Electronics, 704-527-8188
- NC** **Raleigh**  
EMA, 919-781-4139
- OH** **Dayton**  
Del Steffen & Assocs., 513-293-3145
- OH** **Lexington**  
Del Steffen & Assocs., 419-884-2313
- OH** **Westerville**  
Hall-Mark Electronics, 614-891-4555
- OR** **Portland**  
Bell Industries, 503-241-4115
- PA** **Corawalls Heights**  
PMI Sales Office, 215-639-9595
- PA** **Pittsburgh**  
Del Steffen & Assocs., 412-276-7366
- TX** **Dallas**  
PMI Sales Office, 214-341-1742
- TX** **Dallas**  
Pioneer/Harvey Electronics, 214-386-7300
- TX** **Houston**  
PMI Sales Office, 713-481-6460
- TX** **Houston**  
Pioneer/Harvey Electronics, 713-988-5555
- WA** **Bellevue**  
PMI Sales Office, 206-641-6008
- WA** **Redmond**  
Anthem Electronics, 206-881-0850
- Can** **Brampton, Ontario**  
Zentronics, 416-451-9600
- Can** **Calgary, Alberta**  
Future Electronics, Inc., 408-259-6408
- Can** **Calgary, Alberta**  
Zentronics, 403-272-1021
- Can** **Edmonton, Alberta**  
Intek Electronics Ltd., 403-437-2755

# MANUFACTURERS & DISTRIBUTORS DIRECTORY

## Precision Monolithics (Cont)

Can	<b>Nepaan, Ontario</b> Zentronics, 613-226-8840
Can	<b>Rexdale, Ontario</b> Source Elctns. Ltd., 416-675-6235
Can	<b>Richmond, B.C.</b> Zentronics, 604-273-5575
Can	<b>Vancouver, B.C.</b> Future Electronics, inc., 604-438-5545
Can	<b>Waterloo, Ontario</b> Zentronics, 519-884-5700
Can	<b>Winnipeg, Manitoba</b> Zentronics, 204-775-8661
Intl	<b>Australia, Victoria</b> RIFA, TEL: 03-480-1211
Intl	<b>Austria, Vienna</b> Ing. Otto Folger, TEL: 0222/ 43 26 39
Intl	<b>Benelux, Voorburg</b> Bourns (Nederland) B.V., TEL: 070 87 44 00
Intl	<b>Denmark, Ballerup</b> Tage Olsen A/S, TEL: 02-65 81 11
Intl	<b>Eastern Europe, Vienna</b> Eltrans, TEL: 0222/24 71 37
Intl	<b>Finland, Helsinki</b> Insele Oy, TEL: 80/73 57 74
Intl	<b>France, Paris</b> Ohmic S.A., TEL: 01/203 9633
Intl	<b>Germany, Stuttgart</b> Bourns GmbH, TEL: 0711/24 29 36
Intl	<b>Greece, Athens</b> Germanis Co., TEL: 01-821 5825
Intl	<b>India, Bombay</b> Hindco Corp., TEL: 373544
Intl	<b>Israel, Tel Aviv</b> Rapac Elctns. Ltd., TEL: 03-47 71 15
Intl	<b>Italy, Milan</b> Technic S.r.L., TEL: 02/569 57 46
Intl	<b>Japan, Tokyo</b> Nippon PMI Corp., TEL: (03) 234-1411
Intl	<b>Portugal, Lisbon</b> Telectra S.A.R.L., TEL: 19/68 60 72
Intl	<b>Spain, Madrid</b> Hispano Elctrna. S.A., TEL: 01-619-4108
Intl	<b>Sweden, Taaby</b> Bexab Elektronik AB, TEL: 08-768 0560
Intl	<b>Switzerland, Basle</b> Bourns (Schweiz) AG, TEL: 042/33 33 33
Intl	<b>Turkey, Ankara</b> Nol, Nukleer Elektronik Ltd., TEL: 041-30 15 10
Intl	<b>United Kingdom, Middlesex</b> Bourns Electronics, Ltd., TEL: 01/572 6531
Intl	<b>Yugoslavia, Zagreb</b> Jugomineral, TEL: 041-44 1885

### Distributors

AL	<b>Huntsville</b> Hall-Mark Elctns., 205-837-8700
CA	<b>Chatsworth</b> Anthem Elctns., 213-700-1000
CA	<b>Gardena</b> Bell Inds., 213-515-1800
CA	<b>Los Alamitos</b> SEMI-DICE, Inc., 213-594-4631
CA	<b>Reseville</b> Bell Inds., 916-969-3100
CA	<b>San Diego</b> Anthem Elctns., 619-279-5200
CA	<b>San Diego</b> L & S Associates, 619-455-0055
CA	<b>San Jose</b> Anthem Elctns., 408-946-8000
CA	<b>Sunnyvale</b> Bell Inds., 408-734-8570
CA	<b>Sunnyvale</b> Hall-Mark Elctns., 408-773-9990
CA	<b>Tustin</b> Anthem Elctns., 714-730-8000

CO	<b>Denver</b> Hall-Mark Elctns., 303-694-1662
CO	<b>Wheatridge</b> Bell Inds., 303-424-1985
CT	<b>Norwalk</b> Pioneer/Harvey Elctns., 203-853-1515
FL	<b>Boca Raton</b> Lawrence Associates, Inc., 305-368-7373
FL	<b>Clearwater</b> Lawrence Associates, Inc., 813-584-8110
FL	<b>Fort Lauderdale</b> Hall-Mark Elctns., 305-971-9280
FL	<b>Orlando</b> Hall-Mark Elctns., 305-855-4020
IL	<b>Bensenville</b> Hall-Mark Elctns., 312-860-3800
IL	<b>Elk Grove Village</b> Pioneer/Harvey Electronics, 312-437-9680
IN	<b>Indianapolis</b> Pioneer/Harvey Electronics, 317-849-7300
KS	<b>Lenexa</b> Hall-Mark Elctns., 913-888-4747
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<b>ADT</b>	Advanced Digital Technology			<b>OEI</b>	Optical Electronics Inc.
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<b>AMA</b>	American Automation	<b>Heurikon</b>	Heurikon Corp.	<b>Omnibyte</b>	Omnibyte Corp.
<b>AMD</b>	Advanced Micro Devices	<b>Hilevel</b>	Hilevel Technology, Inc.	<b>Onset</b>	Onset Computer Corp.
<b>AMI</b>	American Microsystems, Inc.	<b>Hitachi</b>	Hitachi America, Ltd.		
<b>Amperex</b>	Amperex Electronic Corp.	<b>Holt</b>	Holt Inc.		
<b>Analogic</b>	Analogic	<b>HP</b>	Hewlett-Packard	<b>Panasonic</b>	Panasonic
<b>Analog Sys</b>	Analog Systems	<b>Hughes</b>	Hughes Aircraft, Solid State Products	<b>Pico Design</b>	Pico Design
<b>APC</b>	Applied Micro Circuits	<b>Hybrid Sys</b>	Hybrid Systems	<b>Polycore</b>	Polycore Electronics
<b>Apex</b>	Apex Microtechnology	<b>HyComp</b>	HyComp	<b>Plessey</b>	Plessey Semiconductors
<b>APM</b>	Applied Microsystems Corp.			<b>PMI</b>	Precision Monolithics, Inc.
<b>Appl Sys</b>	Applied Systems Corp.			<b>PragDes</b>	Pragmatic Design Inc.
<b>APT</b>	Applied Microtechnology			<b>Pro-Log</b>	Pro-Log Corp.
<b>Aptek</b>	Aptek Microsystems	<b>ICC</b>	International Cybernetics		
<b>Array Tech</b>	Array Technology	<b>IDT</b>	Integrated Device Technology	<b>Quay</b>	Quay Corp.
<b>AWI</b>	AWI Electronics	<b>IMI</b>	International Microcircuits, Inc.		
		<b>IMP</b>	International Microelectronic Products	<b>Raytheon</b>	Raytheon Semiconductor
			Industrial MicroSystems Inc.	<b>RCA</b>	RCA Solid State Division
<b>Barvon</b>	Barvon Research	<b>IMS</b>	Infosphere	<b>RCI Data</b>	RCI Data
<b>Bedford</b>	Bedford Computer Systems Inc.	<b>Infmos</b>	Infmos	<b>RELMS</b>	Relational Memory Systems
<b>Burr-Brown</b>	Burr-Brown	<b>IntCirEng</b>	Integrated Circuit Engineering	<b>Reticon</b>	Reticon
		<b>IntCirSys</b>	Integrated Circuit Systems	<b>RIFA</b>	RIFA
		<b>IntCompSys</b>	Integrated Computer Systems	<b>Rockwell</b>	Rockwell, Microelectronic Devices
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<b>Cal Devices</b>	California Devices	<b>Interdesign</b>	Interdesign	<b>SBE</b>	SBE, Inc.
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<b>Cherry</b>	Cherry Semiconductor	<b>ITT</b>	ITT Semiconductors	<b>Siemens</b>	Siemens
<b>CIC</b>	Custom Integrated Circuits			<b>Si-Fab</b>	Si-Fab
<b>CirTech</b>	Circuit Technology	<b>Kinetic Sys</b>	Kinetic Systems	<b>Signetics</b>	Signetics
<b>Citel</b>	Citel	<b>Kontron</b>	Kontron Electronics	<b>SOS</b>	SGS Semiconductor
<b>Comlinear</b>	Comlinear Corporation			<b>Sharp</b>	Sharp
<b>CMA</b>	Custom MOS Arrays	<b>Lambda</b>	Lambda Semiconductor	<b>Silicon G</b>	Silicon General
<b>Comark</b>	Comark Corp.	<b>Linear Tech</b>	Linear Technology	<b>Siliconix</b>	Siliconix
<b>Comdial</b>	Comdial Semiconductor	<b>LSI Comp</b>	LSI Computer Systems	<b>Silicon Sys</b>	Silicon Systems Inc.
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<b>Control Sys</b>	Control Systems Microsystems Div.	<b>Matrix</b>	Matrix Corp.	<b>Solarise</b>	Solarise Enterprises
<b>CreMicro</b>	Creative Micro Systems	<b>Matrox</b>	Matrox Electronic Systems	<b>Solitron</b>	Solitron Devices
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<b>Cybersys</b>	Cybersystems	<b>Micro Innov</b>	Micro Innovators	<b>STC</b>	Storage Technology Corp.
<b>Cybertek</b>	Cybertek Inc.	<b>Micropac</b>	Micropac Industries	<b>STD</b>	STD Microsystems
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		<b>MillerTron</b>	MillerTronics	<b>Synertek</b>	Synertek
		<b>Miller</b>	Miller Technology	<b>Sys Innov</b>	Systems Innovations
		<b>Mitel</b>	Mitel Semiconductor		
		<b>Mitsubishi</b>	Mitsubishi Electronics	<b>Tau Zero</b>	Tau Zero Inc.
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				<b>Tektronix</b>	Tektronix
		<b>Mostek</b>	Monolithic Systems Corp.	<b>Teledyne C</b>	Teledyne Crystalonics
		<b>Motorola</b>	Motorola	<b>Teledyne P</b>	Teledyne Philbrick
		<b>MRC</b>	Motorola Semiconductor	<b>Teledyne S</b>	Teledyne Semiconductor
		<b>Murray</b>	Murray Consulting	<b>Telefunken</b>	Telefunken
		<b>Monosil</b>		<b>Telmos</b>	Telmos
				<b>Teltone</b>	Teltone Corporation
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				<b>Universal</b>	Universal Semiconductor, Inc.
				<b>Varix</b>	Varix Corp.
				<b>VLSI Design</b>	VLSI Design Associates
				<b>VTI</b>	VLSI Technology, Inc.
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				<b>Weitek</b>	Weitek Corporation
				<b>Western</b>	Western Digital
				<b>Wintek</b>	Wintek Corp.
				<b>Xicor</b>	Xicor, Inc.
				<b>Xycom</b>	Xycom
				<b>Zendex</b>	Zendex Corp.
				<b>Zilog</b>	Zilog
				<b>ZyMOS</b>	ZyMOS Corporation
				<b>Zytrax</b>	Zytrax Corp.
<b>Fairchild</b>	Fairchild	<b>National</b>	National Semiconductor		
<b>Ferranti</b>	Ferranti Electric	<b>NCM</b>	NCM Corp.		
<b>Force</b>	Force Computers	<b>NCR</b>	NCR Corp., Microelectronics Division		
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<b>Fujitsu</b>	Fujitsu Microelectronics, Inc.	<b>Nitron</b>	Nitron		



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		intl	<b>Peru, Lima</b> Deltron S.A.
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		intl	<b>Philippines, Metro Manila</b> Semitronics
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		intl	<b>Singapore</b> Edward Eu & Co., Ltd.
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- intl** **Thailand, Bangkok**  
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- intl** **Trinidad, Port-of-Spain**  
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Elkose GmbH, TEL: 07141/4871
- intl** **West Germany, Nurnberg**  
Beck GmbH & Co., Elektronik Bauelemente KG, TEL: 0911/34961-66
- intl** **West Germany, Putzbrunn bei Muenchen**  
Sasco GmbH, TEL: 089/46111
- intl** **West Germany, Quickborn**  
Alfred Neye Enatechnik GmbH, TEL: 04106/ 6121
- intl** **West Germany, Quickborn**  
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Da Costa and Musson Ltds., TEL: 608-50
- intl** **Yugoslavia, Ljubljana**  
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- intl** **Zimbabwe, Harare**  
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**RCI/Data**

RCI/Data  
1992 Lakewood Road  
Toms River, New Jersey 08753  
201-341-5592

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Relational Memory Systems (RELMS)  
1180 Miraloma Way  
Sunnyvale, California 94086  
408-732-5520  
TWX: 910-379-0014

**Distributors**

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- intl** **India, Bombay**  
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- intl** **Israel, Tel-Aviv**  
Vectronics Ltd., TEL: 234424, 228472, 246312
- intl** **Japan, Tokyo**  
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- intl** **South Africa, Pinetown**  
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Interface S.A., TEL: (3) 301 7851
- intl** **Sweden, Stockholm**  
A.B. Gosta Backstrom, TEL: (46) 08/54 1080
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Rapid Recall, Ltd., TEL: (6285) 24961

**EG&G Reticon**

EG&G Reticon Corporation  
345 Potrero Avenue  
Sunnyvale, California 94086-9930  
408-738-4266  
TWX: 910-339-9343

**Specific product information:**

- John Skuria
- Application engineering:**
- Cameras ..... John Skuria
- Analog ..... Tim Small
- Image Sensor .....

**Literature:**

Bobbie Stern

**Price and delivery:**

Bobbie Stern

**Follow-up an order:**

Bobbie Stern

**All other information:**

Alex Findlay

**Sales Office & Representatives**

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- intl** **Austria, Wien**  
Othmar Lackner, TEL: (0222) 53-11-85

- intl** **Belgium, Bruxelles**  
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- intl** **Denmark, Copenhagen**  
Scandinavian Supply, TEL: (01) 83 50 90
- intl** **Finland, Espoo**  
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- intl** **France, Sevres**  
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- intl** **Germany, Muenchen**  
Opteltec, Div. Neumuller GmbH, TEL: (089) 611 81
- intl** **Germany, Munich**  
EG & G Reticon, TEL: 089 91 80 60
- intl** **Israel, Tel Aviv**  
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- intl** **Italy, Milano**  
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- intl** **Italy, Roma**  
Eledra 3S SPA, TEL: (06)812 73 24
- intl** **Japan, Tokyo**  
Tokyo Electron Ltd., TEL: 03-343-4411
- intl** **New Zealand, Auckland**  
Professional Elctns. Ltd., TEL: 493-029
- intl** **Norway, Oslo**  
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- intl** **South Africa, Pretoria**  
Electronic Bldg., TEL: 78-9221/6
- intl** **Sweden, Stockholm**  
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- intl** **Switzerland, Zurich**  
Industrade AG, TEL: (01) 363 22 30
- intl** **United Kingdom, Berks**  
EG & G Reticon, TEL: (0734) 788666

**Riehl Time**

Riehl Time Corporation  
53 South Jefferson Road  
Whippany, New Jersey 07981  
201-887-4141

**RIFA**

RIFA AB  
11 Bromma  
Sweden,

**Rockwell International**

Rockwell International  
Semiconductor Products Division  
4311 Jamboree Rd., P.O. Box C  
Newport Beach, California 92660  
714-833-4700  
TWX: 910-591-1654

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- intl** **Israel, Givatayim**  
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- intl** **Israel, Tel Aviv**  
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- intl** **Japan, Osaka**  
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- intl** **Japan, Tokyo**  
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Kyokuto Boeki Kaisha, Ltd., TEL: (03) 244-3803
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- Int'l **Peoples Republic China, Beijing**  
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- Int'l **Thailand, Bangkok**  
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- Int'l **France, Hertzheim**  
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- Int'l **Greece, Thessaloniki**  
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- Int'l **Italy, Milano**  
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Comelta, TEL: 300-7712
- Int'l **Spain, Madrid**  
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- Int'l **United Kingdom, England**  
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- Int'l **West Germany, Munich**  
Astronic GmbH, TEL: 30 40 11

S MOS Systems

S MOS Systems, Inc.  
50 West Brokaw Road, Suite 7  
San Jose, California 95110  
408-288-8928

Sanyo

Sanyo Semiconductor Corporation  
7 Pearl Court  
Allendale, New Jersey 07401  
201-825-8080  
TELEX: 135138 SANYOSEMI ALNJ

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SBE, Inc.  
4700 San Pablo Avenue  
Emeryville, California 94608  
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SEEQ Technology

SEEQ Technology, Incorporated  
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San Jose, California 95131  
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- Int'l **Switzerland, Zurich**  
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- Int'l **West Germany, Hamburg**  
Microscan GmbH, TEL: 406 30 5067
- Int'l **West Germany, Martinsried (Munich)**  
Atlantik Elektronik GmbH, TEL: (089) 857 2086

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AZ	Tempe Marshall Industries, 602-968-6181
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CA	Ventura Bond Electro Sales, 805-653-6487
CO	Denver Marshall Industries, 303-427-1818
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CT	Wallingford Sterling Electronics, 203-265-9535
FL	Orlando Chip Supply Inc., 305-275-3810
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IL	Arlington Heights Reptron Electronics, 312-593-7070
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IN	Carmel Altex Electronics Corp., 317-848-1323, 800- 692-6039
MD	Gaithersburg Marshall Industries, 301-840-9450
MA	Burlington Marshall Industries, 617-272-8200
MA	Mendon ADD Elctns. Corp., 617-478-4200
MA	Waltham Sterling Electronics, 617-894-6200, 800-225- 8743
MI	Livonia Marshall Industries, 313-525-5850

MI	Livonia Reptron Electronics, Inc., 313-525-2700
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NM	Albuquerque Bell Inds., 505-292-2700
NY	East Syracuse ADD Elctns. Corp., 315-437-0300
NY	Endwell Marshall Industries, 607-754-1570
NY	Rochester Marshall Industries, 716-235-7620
OH	Columbus Reptron Electronics Inc., 614-436-6675
OH	Dayton Bell Inds., 513-434-8231
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OR	Lake Oswego Bell Inds., 503-241-4115
TX	Austin ACT Austin, 512-452-5254
TX	Austin Marshall Industries, 512-458-5654
TX	Dallas Active Comp. Technology, 214-980-1888
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TX	Houston Marshall Industries, 713-789-6600
UT	Salt Lake City ACT Utah, 801-487-8131
UT	Salt Lake City Bell Inds., 801-972-6969
WA	Bellevue Bell Inds., 206-747-1515
WA	Bellevue Marshall Industries, 206-747-9100
WI	Waukesha Bell Inds., 414-784-0235
Intl	Austria, Vienna Othmar Lackner, TEL: 0222 531185
Intl	Belgium, Brussels Auriema N.V., TEL: (02) 52362 95
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Intl	France, Antony Cedex Almex SA, TEL: (1) 666 21 12
Intl	Israel, Ramat-Gan M.R.B.D. Ltd., TEL: 03-736479, 732624
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Intl	West Germany, Hamburg Microscan GmbH, TEL: 406 30 5067
Intl	West Germany, Martinsried (Munich) Atlantik Elektronik GmbH, TEL: (089) 857 2086
Intl	West Germany, Viersen (Dusseldorf) Mostron GmbH, TEL: 02162 17024

SGS Semiconductor	
SGS Semiconductor Corporation 1000 E. Bell Road Phoenix, Arizona 85022 602-867-6100 TELEX: 249976	
<b>Specific product information:</b> Product Marketing ..... Giovanni Seghezze	
<b>Application engineering:</b> Applications ..... Tom Hopkins	
<b>Literature:</b> Sales Department ..... Carole Walker	
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Intl	Italy, Roma Sales Office, TEL: 06-8392848
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**Follow-up on order:**  
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### Application engineering:

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Mospower FET's ..... ext. 2586  
Micropower Linear ..... ext. 2586  
Switches ..... ext. 2022

**Literature:**  
Publications ..... ext. 2066

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**Follow-up an order:**  
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EBV Elektronik GmbH, TEL: 089-61105-1

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**Intl Spain, Madrid**  
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**Intl Spain, Madrid**  
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**Intl Sweden, Solna**  
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Macro-Tek, TEL: 01-428-3265

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**WI Milwaukee**  
Marsh Elctns., Inc., 414-475-6000

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### Silicon Systems

Silicon Systems Inc.  
14351 Myford Road  
Tustin, California 92680  
714-731-7110  
TWX: 910-595-2809

**Specific product information:**  
Larry Pepper

**Literature:**  
Larry Pepper

**Price and delivery:**  
Larry Pepper

### Siltronics

Siltronics Ltd.  
436 Highway 7  
Kanata, Ontario K2L 1T9  
613-836-5003

### Solarise Enterprises

Solarise Enterprises Inc.  
10080 North Wolfe Road, Suite SW3-180  
Cupertino, California 95014  
408-446-9343

### Solid State Micro Technology for Music, Inc.

Solid State Micro Technology for Music, Inc.  
2076B Walsh Avenue  
Santa Clara, California 95050  
408-727-0917  
TELEX: 171189

**Specific product information:**  
Dan Parks

**Application engineering:**  
Ron Dow

**Literature:**  
Receptionist

**Price and delivery:**  
Dan Parks

**Follow-up on order:**  
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SIGEP S.R.L., TEL: (02) 8258053
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I.C. Inventor, TEL: 08-756 50 24

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Svensk Teleindustri, TEL: 08 89 04 35
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Eibatex AG, TEL: 056 26 56 41
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Kurt Hirt AG, TEL: 01/302 21 21
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Semiconductor Division  
8808 Balboa Avenue  
San Diego, California 92123  
714-278-8780

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- Intl** **Brazil, Sao Paulo**  
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- Intl** **Finland, Helsinki**  
Oy Fintronic A-b, TEL: 601155
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Nordisk Elektronik (Norge) AS, TEL: 02/553893
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- Intl** **West Germany, Stuttgart**  
Bitronic GmbH, TEL: 07144 3201

### Sprague Electric

Sprague Electric Company  
Semiconductor Division Integrated Circuit Operations  
115 Northeast Cutoff  
Worcester, Massachusetts 01606  
617-853-5000  
TWX: 710-340-6304

**Sprague Electric (Cont)**

**Specific product information:**  
 Interface (peripheral power)—Except Military ..... P. Emerald  
 Interface (peripheral power)—Military ..... W. Sullivan  
 Linear-Radio, Audio, SMPS ..... R. Mailloux  
 Optical & Telecommunications ..... S. Dumery

**Sprague Electric**

**Sprague Electric Company**  
 Semiconductor Division Discrete Components Operation  
 70 Pembroke Road, P.O. Box 917A  
 Concord, New Hampshire 03301  
 603-224-1961  
 TWX: 710-361-1495

**Specific product information:**  
 Hall-Effect I.C.s ..... T. Yasewicz

**Application engineering:**  
 P. Boehm

**Literature:**  
 647 Marshall Street, North Adams, Massachusetts 01247  
 (413) 664-4411, Ext. 2107

**Price and delivery:**  
 T. Yasewicz

**Place an order:**  
 T. Yasewicz

**Follow-up an order:**  
 T. Yasewicz

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Sprague Elec. Co., 602-244-0154, 602-831-6762
- CA** Beanta  
R. David Miner Inc., 619-267-3900
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 35 Marcus Boulevard  
 Hauppauge, New York 11787  
 516-273-3100  
 TWX: 510-227-8898

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**Application engineering:**  
 Vince Rende

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 Camille Fraschilla

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408-988-0725  
TELEX: 172931

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Synertek Inc.  
P.O. Box 552 (95052), 3001 Stander Way, M/S 34  
Santa Clara, California 95054  
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TWX: 910-338-0135

Specific product information:  
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Teledyne Japan K.K., TEL: 03-4038141
- intl Netherlands, A.J. Capelle, A.B. IJssel**  
Alcom Electronics, TEL: 519533
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Svensk Tele Industri AB, TEL: 380320
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**Teledyne Philbrick**

Teledyne Philbrick  
Allied Drive at Route 128  
Dedham, Massachusetts 02026  
617-329-1600  
TWX: 710-348-6726

**Specific product information:**

- Sales Department
- Application engineering:**  
Sales Department
- Literature:**  
Publications Dept.
- Price and delivery:**  
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- Follow-up an order:**  
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- All other information:**  
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**Sales Office & Representatives**

- intl France, Levallois Perret**  
Teledyne Philbrick France, TEL: (1) 757-19-40
- intl Japan, Osaka**  
Teledyne Philbrick Japan, TEL: (06) 365 0431
- intl Japan, Tokyo**  
Teledyne Philbrick Japan, TEL: 403-8921
- intl United Kingdom, Middlesex**  
Teledyne Philbrick UK, TEL: (01) 897-2501
- intl West Germany, Bremen**  
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- intl West Germany, Munich**  
Teledyne Philbrick Germany, TEL: 089/7915773
- intl West Germany, Wiesbaden**  
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- intl Taiwan, Taipei**  
Indl. Elctrn., TEL: (02) 771-4257

**Teledyne Semiconductor**

Teledyne Semiconductor  
1300 Terra Bella Avenue  
Mountain View, California 94043  
415-968-9241  
TWX: 910-379-6494

**Specific product information:**

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ext. 238
- Literature:**  
ext. 451 & 239
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ext. 304 & 441
- Follow-up an order:**  
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Intl	<b>Finland, Helsinki</b> Fintronic A/B, TEL: 90-6926022	Intl	<b>Finland, Helsinki</b> Fintronic A/B, TEL: 90-6926022
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Semitron Elektronik GmbH, TEL: 07742/7011
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Metronix GmbH, TEL: 089/6114063

**Telefunken**

AEG-Telefunken Corporation  
P.O. Box 3800, Route 22 at Orr Drive  
Somerville, New Jersey 08876  
201-722-9800  
TELEX: 833409

**Telephonics LSI**

Telephonics LSI  
790 Park Avenue  
Huntington, New York 11743  
516-549-6300  
TWX: 510-226-6983

**Telmos**

Telmos Inc.  
740 Kifer Road  
Sunnyvale, California 94086  
408-732-4882  
TWX: 910-339-9623

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**Teltone**

Teltone Corporation  
10801-120th Avenue NE  
Kirkland, Washington 98033  
206-827-9626  
TWX: 910-449-2862  
**Specific product information:**  
Mike Arndt  
**Application engineering:**  
Jim Carefoot  
**Literature:**  
Sales Support Department  
**Price and delivery:**  
Order Entry Department  
**Follow-up an order:**  
Order Entry Department  
**All other information:**  
Mike Arndt

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**Texas Instruments, Inc.**

Texas Instruments, Inc.  
Semiconductor Group  
P.O. Box 401560  
Dallas, TX 75240  
214-995-2011  
TWX: 910-867-4702

**Sales Office & Representatives**

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**Intl** **Japan, Tokyo**  
Rikei Corp., TEL: Tokyo (03) 345-1411

**Intl** **Korea R.O.K., Gwanak-Ku**  
Duksung Trading Co., TEL: 854-5047

**Intl** **Netherlands, Koperwerf**  
Koning en Hartman Elektrotechniek B.V., TEL: 70-210101

**Intl** **New Zealand, Auckland**  
Professional Elctns., TEL: 493-029, 499-048

**Intl** **Norway, Oslo**  
Neco A/S, TEL: (02) 25-93-10

**Intl** **Singapore, Singapore**  
Dynamar International Ltd., TEL: 7476188

**Intl** **South Africa, Capetown**  
Electrolink (Pty) Ltd., TEL: 215350

**Intl** **Spain, Barcelona**  
Sisteco S.A., TEL: 321-7347

**Intl** **Sweden, Solna**  
AB Betoma, TEL: 08-820280

**Intl** **Taiwan, Taipei**  
Dynamar International Ltd., TEL: 235-1139

**Intl** **United Kingdom, Kent**  
Candy Elctn. Comps., TEL: 0622-70774/5

**Intl** **United Kingdom, Kent**  
House of Power, TEL: 0689 71531/7

**Intl** **United Kingdom, London**  
Unitrode (U.K.) Ltd., TEL: 01-318-1431/4

**Intl** **United Kingdom, Thames, Oxon**  
Thame Comps. Ltd., TEL: 084-421-4561

**Universal**

**Universal Semiconductor Inc.**  
1925 Zanker Road  
San Jose, California 95112  
408-279-2830  
TWX: 910-338-7617  
All other information:  
Sales & Mktg. .... (408) 279-2830

**Sales Office & Representatives**

**AL** **Huntsville**  
Macro Marketing Assocs., 205-883-9630

**AZ** **Tempe**  
EDS, 602-829-9699

**CA** **San Diego**  
De Pierre & Assocs., 619-578-2555

**IN** **Greenfield**  
France Enterprises, 317-462-6098

**MA** **Waltham**  
Betronic, 617-894-8400

**TX** **Austin**  
O.M. Sales, Inc., 512-452-1838

**TX** **Dallas**  
O.M. Sales, Inc., 214-241-3876

**TX** **Houston**  
O.M. Sales, Inc., 713-789-4426

**WA** **Kirkland**  
Electronic Mktg. Assoc., 206-827-9350

**Distributors**

**CA** **Fullerton**  
Groh Associates, 714-738-7105, 213-923-0963

**CA** **Mtn. View**  
Straube Associates, 415-969-6060

**FL** **Melbourne**  
R-Rep Company, 305-725-6993

**NY** **Manhasset**  
WIN-COR Electronics Sales, 516-627-9474

**PA** **Willow Grove**  
Q.E.D. Electronics, 814

**Varix Corporation**

Varix Corporation  
122 Spanish Village, Suite 608  
Dallas, Texas 75248  
214-620-0925

**Specific product information:**

Byron E. Jacobs  
**Application engineering:**  
Douglas Traster

**Literature:**

Amy Henness

**Price and delivery:**  
Respective sales office

**Follow-up an order:**  
Respective sales office

**All other information:**  
Byron E. Jacobs

**VLSI Design Associates**

VLSI Design Associates  
4010 Moorpark Avenue  
San Jose, California 95117  
408-985-7270

**VLSI Technology**

VLSI Technology, Inc. (VTI)  
1101 McKay Drive  
San Jose, California 95131  
408-942-1810  
TELEX: 278-907  
TELEX: 178-151

**Sales Office & Representatives**

- CA **El Segundo**  
Varigon Assocs., 213-322-1120
- CA **Irvine**  
VLSI Sales Office, 714-833-3103
- CA **San Diego**  
Bestronics, Inc., 619-452-5500
- CA **San Jose**  
VLSI Sales Office, 408-942-1810
- CA **Sunnyvale**  
Emerging Technology, 408-720-0240
- CA **Sunnyvale**  
James Inds., Inc., 408-749-1122
- FL **Boca Raton**  
Lawrence Assocs., Inc., 305-368-7373
- FL **Clearwater**  
Lawrence Assocs., Inc., 305-339-2828
- FL **West Melbourne**  
Lawrence Assocs., Inc., 305-724-8294
- IL **Jarvis**  
James Industries, 312-358-8000
- IL **Schaumburg**  
James Inds., Inc., 312-884-8000
- MA **Burlington**  
VLSI Sales Office, 617-229-6555
- MN **Minneapolis**  
Micro Resources, Inc., 612-830-1454
- MN **Minneapolis**  
Micro Resources, Inc., 612-830-1454
- MN **Albuquerque**  
Varigon Associates, 505-344-5503
- NY **Pleasantville**  
ABC Electronics, 914-769-8088
- NY **William Park**  
ABC Electronics, 516-747-6610
- OH **Rocky River**  
Norm Case Assocs., 216-333-0400
- OR **Portland**  
Micro Sales, Inc., 503-297-4726
- TX **Austin**  
Logic 1 Sales, Inc., 512-459-1297

- TX **Dallas**  
VLSI Sales Office, 214-387-8350
- TX **Houston**  
Logic 1 Sales, Inc., 713-462-0641
- TX **Richardson**  
Logic 1 Sales, Inc., 214-234-0765
- WA **Bellevue**  
Micro Sales, Inc., 206-451-0568

**Votrax**

Votrax Division of Federal Screw Works  
500 Stephenson Highway  
Troy, Michigan 48064  
Business Products Group—313-588-2050  
TELEX: 8102324140 VOTRAX TRMI

**Distributors**

- Int'l **Australia, Sydney**  
Dick Smith Electronics Pty., Ltd., TEL: 61 2 888 3200
- Int'l **West Germany, Wiesbaden**  
Dr. Otto Soskuty, TEL: 6121 463822

**Weitek**

Weitek Corporation  
3255 Scott Boulevard, Building 2B  
Santa Clara, California 95054  
408-727-6625  
TELEX: 171596 AAA Com Suvl

**Sales Office & Representatives**

- Int'l **Japan, Tokyo**  
Nippon Imex Corp.

**Western Digital**

Western Digital Corporation  
2445 McCabe Way  
Irvine, California 92714  
714-557-3550  
TWX: 910-595-1139

**Sales Office & Representatives**

- AL **Huntsville**  
Pioneer/Huntsville, 205-837-9300
- AL **Huntsville**  
Rep. Inc., 205-881-9270
- AZ **Scottsdale**  
Summit Sales, 602-998-4850
- CA **Costa City**  
Bestronics, 213-870-9191
- CA **Irvine**  
Bestronics, 714-979-9910
- CA **Los Altos**  
El Repco, 415-962-0660
- CA **Mountain View**  
San Jose District Office, 415-941-0216
- CA **Newport Beach**  
Western Regional Sales Office, 714-851-1221
- CA **San Diego**  
Bestronics, 619-452-5500
- CA **San Diego**  
San Diego District Office, 619-457-1777
- CA **Woodland Hills**  
Bestronics, 213-704-5616, 714-521-6392
- CO **Denver**  
Component Sales, 303-759-1666
- CT **Norwalk**  
Pioneer-Harvey/Norwalk, 203-853-1515

- CT **Ridgefield**  
Phoenix Sales, 203-438-9644
- FL **Altamonte Springs**  
Pioneer/Florida, 305-834-9090
- FL **Casselberry**  
Southern Regional Sales Office, 305-331-4434
- FL **Clearwater**  
Dyne-A-Mark Inc., 813-441-4702
- FL **Ft. Lauderdale**  
Dyne-A-Mark Inc., 305-771-6501
- FL **Ft. Lauderdale**  
Ft. Lauderdale Branch, 305-421-1355
- FL **Ft. Lauderdale**  
Pioneer, Ft. Lauderdale, 305-771-7520
- FL **Maitland**  
Dyne-A-Mark Inc., 305-831-2097
- FL **Palm Bay**  
Dyne-A-Mark Inc., 305-727-1092
- GA **Norcross**  
Pioneer/Atlanta, 404-448-1711
- GA **Tucker**  
Rep. Inc., 404-938-4358
- IL **Des Plaines**  
Janus, 312-298-9330
- IL **Elk Grove Village**  
Pioneer/Chicago, 312-437-9680
- IL **Oakbrook**  
Chicago District Office, 312-655-8781
- IN **Indianapolis**  
Pioneer/Indiana, 317-849-7300
- IA **Davenport**  
Design Solutions, 319-322-4142
- KS **Wichita**  
Design Solutions, 316-529-0114
- KY **Louisville**  
Fred A. Dorsey Assoc., 502-426-3540
- MD **Galthersburg**  
Pioneer/Washington, 301-921-0660
- MD **Severna Park**  
New Era Sales, 301-544-4100
- MA **Lexington**  
Pioneer-Harvey/Lexington, 617-863-1200
- MA **Marblehead**  
Northeast Regional Sales Office, 617-631-6466
- MA **Webera**  
Component Technology, 617-933-5390
- MI **Grosse Pointe Park**  
Greiner Assoc., 313-499-0188
- MI **Livonia**  
Pioneer/Michigan, 313-525-1800
- MN **Bloomington**  
Central Regional Sales Office, 612
- MN **Edina**  
Mel Foster Tech. Sales, 612-941-9790
- MN **Minnetonka**  
Pioneer/Twin Cities, 612-935-5444
- MO **Ballwin**  
Design Solutions, 314-227-7170
- NJ **Montvale**  
New Jersey District Office
- NJ **Pine Brook**  
Pioneer-Harvey/Pine Brook, 201-227-1262
- NY **Binghamton**  
Pioneer-Harvey/Binghamton, 607-748-8211
- NY **Commack**  
E.R.A., 516-543-0510
- NY **Fairport**  
Pioneer-Harvey/Rochester, 716-381-7070
- NY **Hempstead**  
Pioneer/Military, 516-231-9200
- NY **Rochester**  
Ossmann Components, 716-424-4460
- NY **Syracuse**  
Ossmann Components, 315-455-6611
- NY **Woodbury**  
Pioneer-Harvey/Woodbury, 516-921-8700
- NC **Greensboro**  
Greensboro District Office, 919-299-6733
- NC **Greensboro**  
Pioneer/Carolina, 919-273-4441

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**NC Raleigh**  
Rep. Inc., 919-851-3007

**OH Chagrin Falls**  
Midwest Marketing Assoc., 216-247-6655

**OH Cleveland**  
Pioneer/Cleveland, 216-587-3600

**OH Dayton**  
Midwest Marketing Assoc., 513-433-2511

**OH Dayton**  
Pioneer/Dayton, 513-236-9000

**PA Erdosheim**  
Omni Sales, 215-233-4600

**PA Horsham**  
Pioneer Delaware Valley, 215-674-4000

**PA Pittsburgh**  
Pioneer/Pittsburgh, 412-782-2300

**TN Jefferson City**  
Rep. Inc., 615-475-4105

**TX Austin**  
Pioneer/Austin, 512-835-4000

**TX Austin**  
West & Assoc., 512-454-3681

**TX Dallas**  
Dallas Distrit Office, 214-248-6785

**TX Dallas**  
Pioneer/Dallas, 214-386-7300

**TX Dallas**  
West & Associates, 214-248-7060

**TX Houston**  
Pioneer/Houston, 713-988-5555

**TX Houston**  
West & Assoc., 713-777-4108

**UT Salt Lake City**  
Component Sales, 801-466-8623

**WA Renton**  
Components West, 206-271-5252

**WI Milwaukee**  
Janus, 414-476-9104

**Can Downsview, Ontario**  
Weber Elctns., 416-663-5670

**Intl Australia, Sandringham, Victoria**  
Daneva Australia, TEL: 61/03-5985622

**Intl Belgium, Ougree (Liege)**  
MCA Tronix, TEL: 41/362795, 41/362780

**Intl Brazil, Sao Paulo-SP**  
Cosele Ltda., TEL: 230/1733, 230/1994

**Intl Denmark, Kekkedal**  
C-88, TEL: 45/02-244888

**Intl England, Ilford, Essex**  
Pronto Enctns., TEL: 44/01-5546222

**Intl Finland, Espoo**  
Komdel OY, TEL: 358/90-885011

**Intl France, Levallois-Perret**  
Technology Resources, TEL: 33/1-757-3133

**Intl Hong Kong, Quarry Bay**  
White & Allcock, TEL: 852/5-632112-8, 852/5-644312

**Intl Israel, Givataim**  
E.I.M. Intl., TEL: 761423, 767251

**Intl Italy, Milano**  
Comprel S.R.L., TEL: 39/2-6120641

**Intl Japan, Tokyo**  
Panetron Div., TEL: 81/03-343-4411

**Intl Netherlands, Am Utrecht**  
Diode BV, TEL: 31/030-884214

**Intl Norway, Mosbru**  
Hans. h. Schive, TEL: 47/02-785160

**Intl South Africa, Pinowrie**  
South Continental Dev., TEL: 27/(011) 798-2400

**Intl South India, Trivandrum, PIN Kryptonix**

**Intl Spain, Madrid**  
Diode SA., TEL: 455/3686

**Intl Sweden, Vaellingby**  
Teleimport AB, TEL: 46/08-890265

**Intl Switzerland, Baden Dautwil**  
W. Stolz AG, TEL: 41/056-840151

**Intl Taiwan, Taipei**  
Taiwan Automation, TEL: 886/02-77109403

**Intl United Kingdom, Epsom**  
Europe Regional Sales Office, TEL: 44-3727-42178

**Intl West Germany, Munchen**  
Electronics 2000, TEL: 49/089-434061

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- Los Angeles, CA (213) 700-8700
- Orange County, CA (714) 549-8401
- San Diego, CA (619) 292-5693
- Salt Lake City, UT (801) 486-4134
- Denver, CO (303) 740-8300
- Chicago, IL (312) 595-1000
- Boston, MA (617) 935-6611
- Danbury, CT (203) 797-9674
- Melville, NY (516) 454-6400
- Syracuse, NY (315) 652-5000
- Totowa, NJ (201) 785-1830
- Columbia, MD (301) 995-1226
- Atlanta, GA (404) 449-4133
- Clearwater, FL (813) 443-4514

**DIPLOMAT ELECTRONICS CORPORATION**

**Distributors**

**AL Huntsville**  
R.M. Electronics, 205-852-1550

**AZ Phoenix**  
Kierulff Elctns., 602-243-4104

**AZ Phoenix**  
Wyle EMG, 602-249-2232

**AZ Tempe**  
Anthem Elctns., 602-244-0900

**AZ Tempe**  
Bell Inds., 602-966-7800

**CA Chatsworth**  
Anthem Elctns., 213-700-1000

**CA Commerce**  
Kierulff Elctns., 213-725-0325

**CA El Segundo**  
Wyle EMG, 213-322-8100

**CA Irvine**  
Wyle EMG, 714-863-9953

**CA Irvine**  
Wyle EMG, 714-851-9953

**CA Palo Alto**  
Kierulff Elctns., 415-924-8522

**CA Rancho Cordova**  
Wyle EMG, 916-638-5282

**CA San Diego**  
Anthem Elctns., 619-279-5200

**CA San Diego**  
Kierulff Elctns., 619-278-2112

**CA San Diego**  
Wyle EMG, 619-565-9171

**CA San Jose**  
Anthem Elctns., 408-946-8000

**CA Santa Clara**  
Wyle EMG, 408-727-2500

**CA Sunnyvale**  
Bell Inds., 408-734-8570

**CA Tustin**  
Anthem Elctns., 714-730-8000

**CA Tustin**  
Kierulff Elctns., 714-731-5711

**CO Englewood**  
Anthem Electronics, 303-790-4500

**CO Englewood**  
Kierulff Elctns., 303-790-4444

**CO Thornton**  
Wyle EMG, 303-457-9953

**CO Wheatridge**  
Bell Inds., 303-424-1985

**CT Danbury**  
Diplomat Elctns., 203-797-9674

**CT Wallingford**  
Kierulff Elctns., 203-265-1115

**FL Clearwater**  
Diplomat Elctns., 813-443-3514

**FL Ft. Lauderdale**  
Hammond Electronics, 305-973-7103

**FL Ft. Lauderdale**  
Kierulff Electronics, 305-486-4004

**FL Orlando**  
Hammond Elctns., 305-849-6060, 800-432-6720

**FL St. Petersburg**  
Kierulff Elctns., 813-576-1966

**GA Norcross**  
Diplomat Elctns., 404-449-4133

**GA Norcross**  
Kierulff Electronics, 404-447-5252

**IL Bensenville**  
Diplomat Elctns., 312-595-1000

**IL Chicago**  
Bell Inds., 312-982-9210

**IL Elk Grove Village**  
Kierulff Elctns., 312-640-0200

**IL Lombard**  
R.M. Electronics, 312-932-5150

**IN Indianapolis**  
R.M. Electronics, 317-247-9701

**IA Cedar Rapids**  
DEECO Inc., 319-365-7551

**MD Baltimore**  
Kierulff Electronics, 301-247-5020

**MD Columbia**  
Diplomat Elctns., 301-995-1226

**MD Columbia**  
Future Electronics, 301-995-1222

**MD Columbia**  
Time Electronics, 301-995-1668

**MA Billerica**  
Kierulff Elctns., 617-667-8331

**MA Wilmington**  
R.C. Components, 617-657-4310

**MA Woburn**  
Diplomat Elctns., 617-935-6611

**MA Woburn**  
Time-New England, 617-435-8080

**MN Edina**  
Kierulff Elctns., 612-941-7500

**MO Maryland Heights**  
Kierulff Elctns., 314-739-0855

**NJ Bellmawr**  
Mid-Atlantic Elctns., 609-931-5303

**NJ Fairfield**  
Kierulff Elctns., 201-575-6750

**NJ Totowa**  
Diplomat Elctns., 201-785-1830

**NM Albuquerque**  
Alliance Elctns., 505-292-3360

**NY Farmingdale**  
S.A.I., 516-294-2710

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MANUFACTURERS & DIST DIRECTORY

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NY	<b>Liverpool</b> Diplomat Elctns., 315-652-5000	Intl	<b>Finland, Espoo</b> Komdel OY, TEL: 358/90-885011	CA	<b>Woodland Hills</b> Centaur Corporation, 213-704-1655
NY	<b>Melville</b> Diplomat Elctns., 516-454-6400	Intl	<b>France, Levallois-Perret</b> Technology Resources, TEL: 33/1-757-3133	CO	<b>Denver</b> D/Z Assocs., Inc., 303-429-9369
NY	<b>Rochester</b> Rochester Radio Supply, 716-454-7800	Intl	<b>Hong Kong, Quarry Bay</b> White & Allcock, TEL: 852/5-632112-8, 852/5-644312	CT	<b>Waterbury</b> HLM Assocs., 203-753-9894
NC	<b>Greensboro</b> Hammond Electronics, 919-275-6391	Intl	<b>Israel, Givataim</b> E.I.M. Intl., TEL: 761423, 767251	FL	<b>Clearwater</b> Beacon, 813-796-2378
NC	<b>Raleigh</b> Kierulff Elctns., 919-872-4410	Intl	<b>Italy, Milano</b> Compref S.R.L., TEL: 39/2-6120641	FL	<b>Fl. Lauderdale</b> Beacon, 305-491-1054
OH	<b>Bedford Heights</b> Kierulff Elctns., 216-587-6558	Intl	<b>Japan, Tokyo</b> Panetron Div., TEL: 81/03-343-4411	FL	<b>Maitland</b> Beacon, 305-647-3498
OK	<b>Tulsa</b> Kierulff Elctns., 918-252-7537	Intl	<b>Netherlands, Am Utrecht</b> Diode BV, TEL: 31/030-884214	FL	<b>Melbourne</b> Beacon, 305-724-8010
OK	<b>Tulsa</b> Quality Comps., 918-664-8812	Intl	<b>Norway, Nesbru</b> Hans H. Schive, TEL: 47/02-785160	GA	<b>Atlanta</b> Beacon, 404-256-9640
OR	<b>Hillsboro</b> Wyle EMG, 503-640-6000	Intl	<b>South Africa, Pinetown</b> South Continental Dev., TEL: 27/(011) 798-2400	ID	<b>Rethram</b> Vantage
OR	<b>Lake Oswego</b> Bell Inds., 503-241-4115	Intl	<b>South India, Trivandrum, PIN</b> Kyronix	IL	<b>Elk Grove Village</b> Oasis Sales Corp., 312-640-1850
PA	<b>Broomall</b> Time Mid-Atlantic, 215-359-1200	Intl	<b>Spain, Madrid</b> Diode SA., TEL: 455/3696	IL	<b>Schaumburg</b> Xicor Central Sales Office, 312-490-1310
SC	<b>Greenville</b> Hammond Electronics, 803-233-4121	Intl	<b>Sweden, Vaelingby</b> Teleimport AB, TEL: 46/08-890265	IN	<b>Indianapolis</b> Chuck Horn & Assoc., 317-241-9276
TX	<b>Addison</b> Quality Comps., 214-387-4949	Intl	<b>Switzerland, Baden Dattwil</b> W. Stolz AG, TEL: 41/056-840151	IN	<b>Indianapolis</b> SAI Mktg. Corp., 317-241-9276
TX	<b>Austin</b> Active Components Technology, 512-452-5254	Intl	<b>Taiwan, Taipei</b> Taiwan Automation, TEL: 886/02-77109403	IA	<b>Cedar Rapids</b> Chuck Horn & Associates, 319-393-8703
TX	<b>Austin</b> Kierulff Elctns., 512-835-2090	Intl	<b>West Germany, Muenchen</b> Electronics 2000, TEL: 49/089-434061	KS	<b>Overland Park</b> Kecco, Inc., 913-541-8431
TX	<b>Austin</b> Quality Comps., 512-835-0220	<hr/> <b>Wintek</b> <hr/>			
TX	<b>Dallas</b> Active Components Technology, 214-980-1888	<b>Wintek Corporation</b> 1801 South Street Lafayette, Indiana 47904-2993 317-742-8428			
TX	<b>Dallas</b> Kierulff Elctns., 214-343-2400	<hr/> <b>Distributors</b> <hr/>			
TX	<b>Houston</b> Active Components Technology, 713-496-4000	CA	<b>Santa Clara</b> PBC Associates, 408-727-7244	MA	<b>Boston</b> Xicor East Sales Office, 617-542-2944
TX	<b>Houston</b> Kierulff Elctns., 713-530-7030	CO	<b>Parker</b> Western Marketing Associates, 303-841-2788	MA	<b>Wellesley</b> Advanced Design Group, 617-431-7946
TX	<b>Sugarland</b> Quality Comps., 713-491-2255	IL	<b>Downers Grove</b> Physics Equipment, Inc., 312-960-4053	MI	<b>Brighton</b> SAI Mktg. Corp., 313-227-1786
UT	<b>Salt Lake City</b> Bell Inds., 801-972-6969	TX	<b>Houston</b> Dyna Marketing, 713-937-0278	MI	<b>Bloomington</b> John G. Twist Co., 612-835-2120
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UT	<b>Salt Lake City</b> Wyle EMG, 801-974-9955	<hr/> <b>Xicor</b> <hr/>			
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WA	<b>Bellevue</b> Bell Inds., 206-747-1515	<hr/> <b>Sales Office &amp; Representatives</b> <hr/>			
WA	<b>Bellevue</b> Wyle EMG, 206-453-8300	AL	<b>Huntsville</b> Beacon, 205-881-5031	NC	<b>Raleigh</b> Beacon, 919-787-0330
WA	<b>Tukwila</b> Kierulff Elctns., 206-573-4220	AZ	<b>Scottsdale</b> Thom Luke Sales, Inc., 602-941-1901	OH	<b>Centerville</b> SAI Mktg. Corp., 513-435-3181
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Can	<b>Brampton, Ontario</b> Zentronics Inc., 416-451-9600	CA	<b>Los Altos</b> Ewing-Foley, Inc., 415-941-4525	OH	<b>Zanesville</b> SAI Marketing Corp., 614-454-8942
Can	<b>Darval, Quebec</b> Semad Elctns., 514-636-4614	CA	<b>Newport Beach</b> Xicor Southwestern Sales Office, 714-752-8700	OR	<b>Portland</b> Vantage Corporation, 503-227-1369
Can	<b>Downsview, Ontario</b> Semad Elctns., 416-663-5680, 416-663-5650	<hr/> <b>TX</b> <hr/>			
Can	<b>Ottawa, Ontario</b> Semad Elctns., 613-729-6145	TX	<b>Dallas</b> Nova Mktg., Inc., 214-750-6082	TX	<b>Houston</b> Nova Mktg., Inc., 713-988-6082
Intl	<b>Australia, Sandringham, Victoria</b> Daneva Australia, TEL: 61/03-5985622	UT	<b>Salt Lake City</b> D/Z Assocs., Inc., 801-483-1025, 801-486-4233	VA	<b>Huddleston</b> Micro Comp. Inc., 703-297-7614
Intl	<b>Belgium, Ougree (Liege)</b> MCA Tronix, TEL: 41/362795, 41/362780	WA	<b>Bellevue</b> Vantage Corp., 206-455-3460	WI	<b>Brookfield</b> Oasis Sales Corp., 414-782-6660
Intl	<b>Brazil, Sao Paulo-SP</b> Cosele Ltda., TEL: 230/1733, 230/1994				
Intl	<b>Denmark, Koksodal</b> C-88, TEL: 45/02-244888				

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**Xicor (Cont)**

**WI** Brookfield  
Oasis Sales Corp., 414-782-6660

**Can** Kanata, Ontario  
Kaytronics, Inc., 613-592-6606

**Intl** Belgium, Koerbergen  
Microtron, TEL: 32.16.60.05.86

**Intl** Denmark, Havdrup  
Henckel Elektronik, TEL: 45.2.38.57.91

**Intl** Finland, Espoo  
Oy Atomica AB, TEL: 90.423.533

**Intl** India, Rajji Nagar  
Zenith Electronics Bangalore, TEL: 83645

**Intl** Italy, Cassina de' Pecchi  
Dott. Ing. Giuseppe De Mico S.p.A., TEL: 39.2.95.20.551

**Intl** Korea, Seoul  
Kortronics Enterprise, TEL: 82.2.634.4397

**Intl** Netherlands, IJssel  
Alcom Electronics, TEL: 31.10.51.95.33

**Intl** New Zealand, Auckland  
Delphi Industries Ltd., TEL: 64.9.563.259

**Intl** Norway, Oslo  
Eltron A/S, TEL: 47.2.46.28.70

**Intl** Spain, Madrid  
Amitron S.A., TEL: 34.1.248.58.63

**Intl** Switzerland, Baden-Dattwil  
W. Stolz AG, TEL: 41.56.84.01.51

**Intl** Taiwan, Taipei  
Multitech International Corp., TEL: 886.2.769.1225

**Intl** United Kingdom, Oxon  
Micro Call Ltd., TEL: 44.84.421.5405

**Intl** West Germany, Haar Bei Muenchen  
Xicor-International, TEL: 011 49 89463089

**Distributors**

**AL** Huntsville  
Hall-Mark, 205-837-8700

**AL** Huntsville  
Marshall, 205-881-9235

**AL** Huntsville  
RM-Alabama, Inc., 205-852-1550

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Hall-Mark, 602-243-6601

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Marshall, 602-968-6181

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Marshall, 213-999-5001

**CA** El Monte  
Marshall, 213-442-7204

**CA** Irvine  
Marshall, 714-556-6500

**CA** San Diego  
Hall-Mark, 619-268-1201

**CA** San Diego  
Marshall, 619-578-9600

**CA** Sunnyvale  
Hall-Mark, 408-773-9990

**CA** Sunnyvale  
Marshall, 408-732-1100

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Image Electronics, 714-730-0303

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Marshall, 303-427-1818

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Hall-Mark, 303-694-1662

**CT** Wallingford  
Marshall, 203-265-3822

**FL** Ft. Lauderdale  
Hall-Mark, 305-971-9280

**FL** Orlando  
Hall-Mark, 305-855-4020

**FL** Orlando  
Marshall, 305-841-1878

**FL** St. Petersburg  
Hall-Mark, 813-576-8691

**GA** Norcross  
Hall-Mark, 404-447-8000

**GA** Norcross  
Marshall, 404-923-5750

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Hall-Mark, 312-860-3800

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**IL** Schaumburg  
Marshall, 312-490-0735

**IN** Indianapolis  
RM-Illinois, Inc., 317-247-9701

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Hall-Mark, 913-888-4747

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Hall-Mark, 301-796-9300

**MD** Gaithersburg  
Marshall, 301-840-9450

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Marshall, 617-272-8200

**MA** Wilmington  
RC Components, 617-273-1860

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Hall-Mark, 617-935-9777

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RM-Michigan, Inc., 616-531-9300

**MI** Livonia  
Marshall, 313-525-5850

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Hall-Mark, 612-854-3223

**MN** Plymouth  
Marshall, 612-559-2211

**MO** Maryland Hts  
Hall-Mark, 314-291-5350

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Hall-Mark, 215-355-7300, 609-424-7300

**NJ** Fairfield  
Hall-Mark, 201-575-4415

**NJ** Fairfield  
Marshall, 201-882-0320

**NJ** Mt. Laurel  
Marshall, 609-234-9100, 215-627-1920

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Alliance Electronics, 800-545-6288

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Marshall, 516-273-2424

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Marshall, 716-235-7620

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Hall-Mark, 516-737-0600

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Hall-Mark, 919-872-0712

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Hall-Mark/Cincinnati, 513-563-5980

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Marshall, 513-236-8088

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Hall-Mark, 216-473-2907

**OH** Westerville  
Hall-Mark, 614-891-4555

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Hall-Mark, 918-665-3200

**TX** Austin  
Hall-Mark, 512-258-8848

**TX** Austin  
Marshall, 512-458-5654

**TX** Dallas  
Hall-Mark, 214-343-5000

**TX** Dallas  
Marshall, 214-233-5200

**TX** Houston  
Hall-Mark, 713-781-6100

**TX** Houston  
Marshall, 713-789-6600

**WA** Bellevue  
Marshall, 206-747-9100

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**WI** Oak Creek  
Hall-Mark, 414-761-3000

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Zenitronics, 416-451-9600

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Zenitronics, 403-272-1021

**Can** Concord, Ontario  
Kaytronics Ltd., 416-669-2262

**Can** Montreal, Quebec  
Zenitronics, 514-735-5361

**Can** Ottawa, Ontario  
Zenitronics, 613-237-6411

**Can** Surrey, British Columbia  
Kaytronics Ltd., 604-581-7611

**Can** Vancouver, B.C.  
Zenitronics, 604-273-5575

**Can** Ville St. Pierre, Quebec  
Kaytronics Ltd., 514-367-0101

**Can** Waterloo, Ontario  
Zenitronics, 519-884-5700

**Can** Willowdale, Ontario  
Electro Sonic, 416-494-1666

**Can** Winnipeg, Manitoba  
Zenitronics, 204-775-8661

**Intl** Australia, Burwood, Victoria  
R & D Elctns., TEL: 61.3.288.8232

**Intl** Cyprus, Nicosia  
Poly Electronics Ltd. Middle East, TEL: 357.21.61088

**Intl** France, Le Chesnay  
A2M, TEL: 33.3.954.91.13

**Intl** India, Bombay  
Zenith Electronics, TEL: 384214

**Intl** India, Calcutta  
Zenith Electronics, TEL: 228797

**Intl** India, New Delhi  
Zenith Electronics, TEL: 522281

**Intl** Israel, Tel-Aviv  
Telsys Ltd., TEL: 972.3.494881

**Intl** Japan, Tokyo  
Internix, Inc., TEL: 81.3.369.1105

**Intl** New Zealand, Auckland  
Delphi Inds. Ltd., TEL: 64.9.563.259

**Intl** South Africa, Transvaal  
Taltronics (Pty.) Ltd., TEL: 27.11.834.5151

**Intl** Sweden, Sefes  
Naxab, TEL: 46.8.98.51.40

**Intl** United Kingdom, Thame Oxon  
Micro Call Ltd., TEL: 44.84.421-5405

**Intl** West Germany, Gerlingen  
Positron, TEL: 49.07.156.3560

**Xycom**

Xycom  
750 North Maple Road  
Saline, Michigan 48176  
313-429-4971

**Zendex**

Zendex Corporation  
6644 Sierra Lane  
Dublin, California 94566  
415-828-3000

**Zilog**

Zilog, Inc.  
1315 Dell Avenue  
Campbell, California 95008  
408-370-8000  
TWX: 910-338-7621

**Sales Office & Representatives**

**AL** Huntsville  
Electn. Sales, Inc., 205-533-1735

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AZ	<b>Scottsdale</b> Thom Luke Sales, Inc., 602-941-1901	TX	<b>Dallas</b> Nova Marketing, 214-750-6082	CA	<b>Tustin</b> Kierulff Elctrns., 714-731-5711
CA	<b>Campbell</b> Sales & Tech. Ctr. Zilog, Inc., 408-370-8120	TX	<b>Dallas</b> Sales & Tech. Ctr. Zilog, Inc., 214-931-9090	CO	<b>Denver</b> Kierulff Elctrns., 303-371-6500
CA	<b>Capertee</b> Costar, Inc., 408-446-9339	TX	<b>Houston</b> Nova Marketing, 713-988-6082	CO	<b>Englewood</b> Hallmark Electronics, 303-694-1662
CA	<b>Irvine</b> Sales & Tech. Ctr. Zilog, Inc., 714-549-2891	TX	<b>Houston</b> Zilog, Inc., 713-840-5475	CO	<b>Wheatridge</b> Bell Inds., 303-424-1985
CA	<b>San Diego</b> Littlefield & Smith Assoc., 619-455-0055	UT	<b>West Valley</b> Thorson Rocky Mountain, 801-973-7969	CT	<b>Danbury</b> Schweber Elctrns., 203-792-3500
CA	<b>Van Nuys</b> Sales & Tech. Ctr. Zilog, Inc., 213-989-7485	VA	<b>Arlington</b> Zilog, Inc., 703-243-7900	CT	<b>Wallingford</b> Kierulff Elctrns., 203-265-1115
CO	<b>Englewood</b> Thorson Rocky Mountain, 303-779-0666	VA	<b>Lynchburg</b> Glen White Assoc., 804-237-6291	FL	<b>Altamonte Springs</b> Schweber Elctrns., 305-331-7555
FL	<b>Altamonte Springs</b> Dyne-A-Mark Corp., 305-831-2097	WA	<b>Bellevue</b> Sales & Tech. Ctr. Zilog, Inc., 206-454-5597	FL	<b>Fort Lauderdale</b> Hall-Mark Elctrns., 305-971-9280
FL	<b>Clearwater</b> Dyne-A-Mark Corp., 813-441-4702	WI	<b>Waukesha</b> Micro-Tex, Inc., 414-542-5352	FL	<b>Hollywood</b> Schweber Elctrns., 305-927-0511
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FL	<b>Fort Lauderdale</b> Dyne-A-Mark Corp., 305-771-6501	Can	<b>Etobicoke, Ontario</b> R.F.Q., Ltd., 416-626-1445	FL	<b>St. Petersburg</b> Kierulff Elctrns., Inc., 813-576-1966
FL	<b>Palm Bay</b> Dyne-A-Mark Corp., 305-727-0192	Can	<b>Ottawa, Ontario</b> R.F.Q., Inc., 613-820-8445	GA	<b>Norcross</b> Hall-Mark Elctrns., 404-447-8000
GA	<b>Atlanta</b> Sales & Tech. Ctr. Zilog, Inc., 404-451-8425	Can	<b>West Vancouver, British Columbia</b> Blakewood Elctrn. Sys., Inc., 604-926-8000	GA	<b>Norcross</b> Kierulff Electronics, 404-447-5252
GA	<b>Norcross</b> Elctrn. Sales, Inc., 404-448-6554	Int'l	<b>France, Paris La Defense</b> Zilog, Inc., TEL: (1) 334-60-09	GA	<b>Norcross</b> Schweber Elctrns., 404-449-9170
IL	<b>Schaumburg</b> Sales & Tech. Ctr. Zilog, Inc., 312-885-8080	Int'l	<b>Japan, K.K.</b> Zilog/Japan, TEL: (81) (03) 587-0528	IL	<b>Bensenville</b> Hall-Mark Elctrns., 312-660-3600
IN	<b>Fl. Wayne</b> Giesting & Assoc., 219-486-1912	Int'l	<b>United Kingdom, Berkshire</b> Zilog (U.K.) Ltd., TEL: 0628-39200	IL	<b>Elk Grove Village</b> Kierulff Elctrns., 312-640-0200
IN	<b>Indianapolis</b> Giesting & Assoc., 317-263-0005	Int'l	<b>West Germany, Munich</b> Zilog GmbH, TEL: 89-612-6046	IL	<b>Elk Grove Village</b> Schweber Elctrns., 312-364-3750
KS	<b>Overland Park</b> Advance Technical Sales, 913-492-4333	<b>Distributors</b>		IN	<b>Indianapolis</b> Advent Electronics, 317-872-4910
MO	<b>Soverna Park</b> New Era Sales, 301-544-4100	AL	<b>Huntsville</b> Hall-Mark Elctrns., 205-837-8700	IA	<b>Cedar Rapids</b> Advent Electronics, 319-363-0221
MA	<b>Burlington</b> Sales & Tech. Ctr. Zilog, Inc., 617-273-4222	AL	<b>Huntsville</b> Schweber Elctrns., 205-882-2200	IA	<b>Cedar Rapids</b> Schweber Elctrns., 319-373-1417
MA	<b>Lexington</b> Nova Sales, 617-861-1820	AZ	<b>Phoenix</b> Hall-Mark Elctrns., 602-243-6601	KS	<b>Lenexa</b> Hall-Mark Elctrns., 913-888-4747
MI	<b>Coloma</b> Giesting & Assoc., 616-468-4200	AZ	<b>Phoenix</b> Kierulff Elctrns., 602-243-4101	KS	<b>Overland Park</b> Schweber Elctrns., 913-492-2921
MI	<b>Novi</b> Giesting & Assoc., 313-348-3811	AZ	<b>Scottsdale</b> Western Microtechnology, 602-948-4240	ND	<b>Baltimora</b> Hall-Mark Elctrns., 301-796-9300
MN	<b>Minneapolis</b> H.M.R., Inc., 612-831-7400	AZ	<b>Tempe</b> Anthem Elctrns., 602-244-0900	ND	<b>Getthersburg</b> Schweber Elctrns., 301-840-5900
MO	<b>St. Louis</b> Advance Technical Sales, 314-227-4448	CA	<b>Canoga Park</b> Schweber Elctrns., 213-999-4702	MA	<b>Bedford</b> Schweber Elctrns., 617-275-5100
NJ	<b>Cedar Knolls</b> Sales & Tech. Ctr. Zilog, Inc., 201-540-1671	CA	<b>Chatsworth</b> Anthem Elctrns., 213-700-1000	MA	<b>Billerica</b> Kierulff Elctrns., 617-667-8331
NJ	<b>Clester</b> Vera Associates, 201-768-6100	CA	<b>Capertee</b> Western Microtechnology, 408-725-1660	MA	<b>Billerica</b> Kierulff Electronics, 617-667-8331
NJ	<b>Poenssaken</b> Vantage Sales, 609-663-6660	CA	<b>Irvine</b> Schweber Elctrns., 714-556-3880	MA	<b>Burlington</b> Lionex Corp., 617-272-9400
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NY	<b>E. Syracuse</b> L-Mar Assocs., Inc., 315-437-7779	CA	<b>Palo Alto</b> Kierulff Elctrns., 415-968-6292	MI	<b>Farmington Hills</b> Advent Electronics, 313-447-1650
NY	<b>New York</b> Zilog, Inc., 212-398-4497	CA	<b>Rededa</b> JAN Devices, Inc., 213-708-1100	MI	<b>Livonia</b> Schweber Elctrns., 313-525-8100
NY	<b>Rechester</b> L-Mar Assocs., Inc., 716-323-1000	CA	<b>Sacramento</b> Schweber Elctrns., 916-929-9732	MI	<b>Livonia</b> Schweber Electronics, 313-525-8100
NC	<b>Raleigh</b> Glen White Assoc., 919-787-7016	CA	<b>San Diego</b> Anthem Elctrns., 714-279-5200	MN	<b>Bloomington</b> Hall-Mark Elctrns., 612-854-3223
OH	<b>Cincinnati</b> Giesting & Assoc., 513-521-8800	CA	<b>San Diego</b> Kierulff Elctrns., 619-278-2112	MN	<b>Eden Prairie</b> Schweber Elctrns., 612-941-5280
OH	<b>Dayton</b> Giesting & Assoc., 513-293-4044	CA	<b>San Francisco</b> Kanematsu-Gosho (USA) Inc./Hong Kong & Japan, 415-788-3800	MN	<b>Edina</b> Kierulff Elctrns., 612-941-7500
OH	<b>Monter</b> Giesting & Assoc., 216-942-3407	CA	<b>San Jose</b> Anthem Elctrns., 408-946-8000	MO	<b>Earth City</b> Schweber Elctrns., 314-739-0526
OH	<b>Woodmere</b> Sales & Tech. Ctr. Zilog, Inc., 216-831-7040	CA	<b>Santa Clara</b> Schweber Elctrns., 408-748-4700	MO	<b>Maryland Heights</b> Hall-Mark Elctrns., 314-291-5350
TN	<b>Gray</b> Glen White Assoc., 615-477-8850	CA	<b>Sunnyvale</b> Multitech Elctrns. Inc./Taiwan, 408-773-8400	MO	<b>Maryland Heights</b> Kierulff Elctrns., 314-739-0855
TX	<b>Austin</b> Zilog, Inc., 512-453-3216	CA	<b>Tustin</b> Anthem Electronics, 714-730-8000	NH	<b>Manchester</b> Schweber Electronics, 603-625-2250



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<b>NJ</b>	<b>Fairfield</b> Schweber Elctns., 201-227-7880	<b>Can</b>	<b>Dorval, Quebec</b> Semad, 514-636-4614
<b>NM</b>	<b>Albuquerque</b> Bell Inds., 505-292-2700	<b>Can</b>	<b>Downsview, Ontario</b> Future Elctns., 416-663-5563
<b>NM</b>	<b>Albuquerque</b> International Electronics, 505-345-8127	<b>Can</b>	<b>Downsview, Ontario</b> Semad, 416-663-5680
<b>NY</b>	<b>Hempstead</b> Lionex Corp., 516-273-1660	<b>Can</b>	<b>Edmonton, Alberta</b> R.A.E. Industrial Electronics, Ltd., 403-451-4001
<b>NY</b>	<b>Rechester</b> Schweber Elctns., 716-424-2222	<b>Can</b>	<b>Montreal, Quebec</b> Future Elctns., 514-694-7710
<b>NY</b>	<b>Westbury</b> Schweber Elctns., 516-334-7474	<b>Can</b>	<b>Ottawa, Ontario</b> Future Elctns., 613-820-8313
<b>NC</b>	<b>Greensboro</b> Kierulff Elctns., 919-852-9440	<b>Can</b>	<b>Ottawa, Ontario</b> Semad, 613-722-6571
<b>NC</b>	<b>Raleigh</b> Hall-Mark Elctns., 919-872-0712	<b>Can</b>	<b>Vancouver, British Columbia</b> Future Elctns., 604-438-5545
<b>NC</b>	<b>Raleigh</b> Schweber Elctns., 919-876-8000	<b>Intl</b>	<b>Australia, Adelaide, S.A.</b> Protronics Pty. Ltd. (S.A.), TEL: (61) (08) 212-3111
<b>OH</b>	<b>Beachwood</b> Schweber Elctns., 216-464-2970	<b>Intl</b>	<b>Australia, Burwood, Victoria</b> R & D Electronics, TEL: 011-61-3-2999232
<b>OH</b>	<b>Cleveland</b> Kierulff Elctns., 216-587-6558	<b>Intl</b>	<b>Australia, Campdown</b> George Brown & Co. Pty., Ltd., TEL: (61) (02) 519-5855
<b>OH</b>	<b>Dayton</b> Schweber Elctns., 513-439-1800	<b>Intl</b>	<b>Australia, East Perth</b> Simply Software, TEL: 011-61-9-3259376
<b>OH</b>	<b>Highland Heights</b> Hall-Mark Elctns., 216-473-2907	<b>Intl</b>	<b>Austria Vienna</b> Kontron Elektronik, TEL: 0222-670631
<b>OH</b>	<b>Westerville</b> Hall-Mark Elctns., 614-891-4555	<b>Intl</b>	<b>Belgium, Merelbeke</b> Heliagraph, TEL: 91-301314
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<b>OK</b>	<b>Tulsa</b> Schweber Elctns., 918-622-8000	<b>Intl</b>	<b>Denmark, Naerum</b> SC Metric A/S, TEL: 2-804200
<b>OR</b>	<b>Portland</b> Kierulff Elctns., 503-641-9150	<b>Intl</b>	<b>Finland, Espoo</b> Finn Metric oy, TEL: 0-42-39-11
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<b>PA</b>	<b>Pittsburgh</b> Schweber Elctns., 412-782-1600	<b>Intl</b>	<b>France, Antony</b> Almex, TEL: 666 2112
<b>TX</b>	<b>Austin</b> Hall-Mark Elctns., 512-258-8848	<b>Intl</b>	<b>France, Asnelles Cedex</b> Alfatronic, TEL: 791-44-44
<b>TX</b>	<b>Austin</b> Kierulff Elctns., 512-835-2090	<b>Intl</b>	<b>France, Le Chesnay</b> A2M, TEL: 954-9113
<b>TX</b>	<b>Austin</b> Schweber Elctns., 512-458-8253	<b>Intl</b>	<b>Greece, Athens</b> Microsystems O.E., TEL: 1-346-9002
<b>TX</b>	<b>Dallas</b> Hall-Mark Elctns., 214-343-5000	<b>Intl</b>	<b>Hong Kong</b> Electronic Scientific Eng. Ltd., TEL: 852-05-487007
<b>TX</b>	<b>Dallas</b> Kierulff Elctns., 214-343-2400	<b>Intl</b>	<b>Hong Kong, GPO</b> Electronic Scientific Eng Ltd., TEL: (852) (05) 487007
<b>TX</b>	<b>Dallas</b> Schweber Elctns., 214-661-5010	<b>Intl</b>	<b>India, Sunnyvale, CA</b> Micro Aids International, TEL: 408-733-0695
<b>TX</b>	<b>El Paso</b> International Electronics, 915-778-9761	<b>Intl</b>	<b>India, Trivandrum</b> Kryonix, TEL: 63805
<b>TX</b>	<b>Houston</b> Hall-Mark Elctns., 713-781-6100	<b>Intl</b>	<b>Israel, Ramat Hasharon</b> R.D.T., TEL: 483211-5
<b>TX</b>	<b>Houston</b> Kierulff Elctns., 713-530-7030	<b>Intl</b>	<b>Italy, Milan</b> De Mico SpA, TEL: 02-9520551
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<b>WI</b>	<b>Waukesha</b> Kierulff Elctns., 414-784-8160		
<b>Can</b>	<b>Burnaby, B.C.</b> R.A.E. Industrial Electronics, Ltd., 604-291-8866		
		<b>Intl</b>	<b>Netherlands, Zoetermeer</b> Tekelec Airtronic, TEL: 79-310100
		<b>Intl</b>	<b>New Zealand, Auckland</b> Micro Computer Elctns. Co., Ltd., TEL: (64) (09) 504-774
		<b>Intl</b>	<b>Norway, Oslo</b> Metric A/S, TEL: 2-282624
		<b>Intl</b>	<b>People's Republic of China, San Francisco, CA</b> Kanematsu-Gosho Inc., TEL: 415-788-3800
		<b>Intl</b>	<b>Peru, Lima</b> ITT del Peru, TEL: (51) (14) 401715
		<b>Intl</b>	<b>Portugal, Lisbon</b> Soc. Com. Crocker Delaforce & Co. Ltd., TEL: 68-01-41424344
		<b>Intl</b>	<b>South Africa, Transvaal</b> Radiokom Pty., Ltd., TEL: (001) 789-1400/5
		<b>Intl</b>	<b>Spain, Barcelona</b> Unitronics S.A., TEL: 1-242-5204
		<b>Intl</b>	<b>Sweden, Solna</b> Scandia Metric AB, TEL: 8-820 400
		<b>Intl</b>	<b>Sweden, Stockholm</b> AB Nordquist & Berg, TEL: 46-8-69-04-00
		<b>Intl</b>	<b>Switzerland, Baden-Daetwil</b> W. Stotz AG, TEL: 056 84 01 51
		<b>Intl</b>	<b>Switzerland, Rogensdorf</b> V. Moor AG, TEL: 01-8406644
		<b>Intl</b>	<b>Taiwan, Taipei</b> Multitech Intl. Corp., TEL: (886) (02) 769-1225
		<b>Intl</b>	<b>Turkey, Istanbul</b> Electro Akustik Sanayi, TEL: (90) 11-48-50-77
		<b>Intl</b>	<b>United Kingdom, Bar Hill, Cambridge</b> Hi-Tek Distribution, TEL: 954-81996
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		<b>Intl</b>	<b>United Kingdom, Slough, Berkshire</b> Crellon Elctns. Ltd., TEL: 6286-4434/4300
		<b>Intl</b>	<b>West Germany, Echting</b> Kontron Elektronik GmbH, TEL: 089-319011
		<b>Intl</b>	<b>West Germany, Uterkaching</b> E.B.V. Elektronik, TEL: 89-611051

## ZyMOS

ZyMOS Inc.  
P.O. Box 62379  
Sunnyvale, California 94088  
408-730-8800

## Zytrex Corporation

Zytrex Corporation  
224 North Wolfe Road  
Sunnyvale, California 94086  
408-733-3973